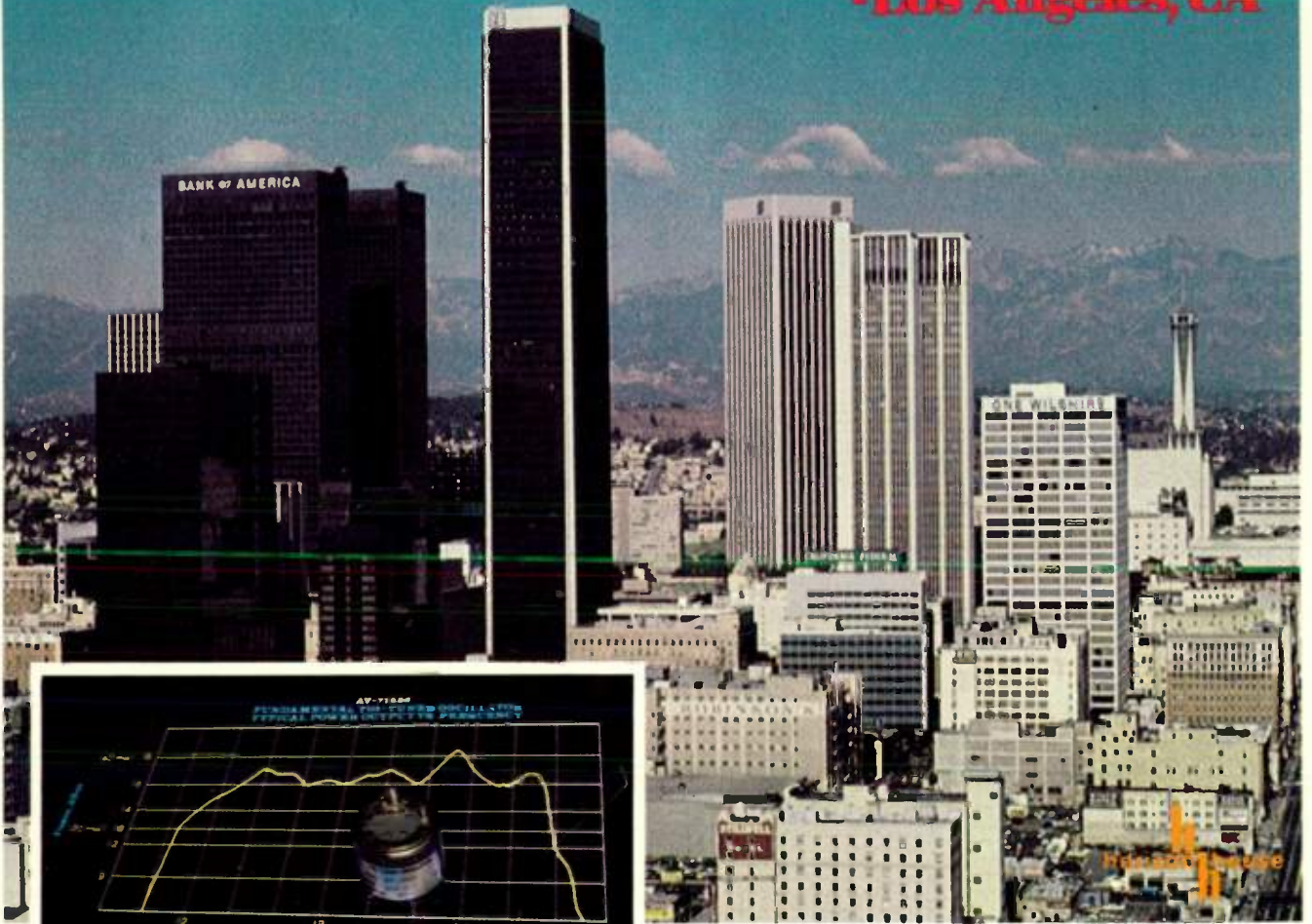




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MIT-S, AP-S, URSI Programs
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Super Component Improves Airborne IFF Interrogator Performance

Complete transmit and receiver functions are provided in a single microwave integrated circuit with significant reduction in total insertion loss, size, and weight, compared to discrete components.

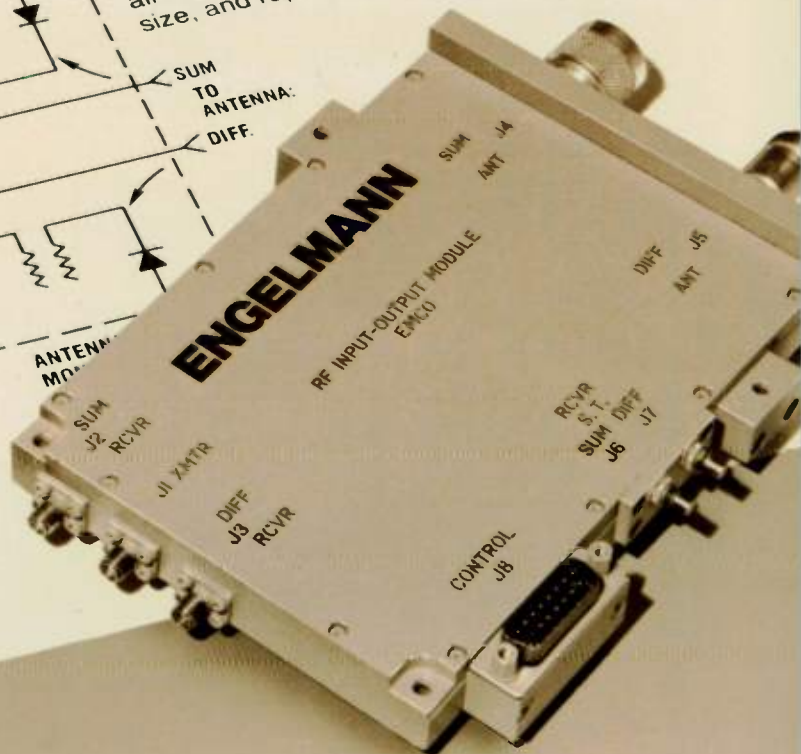
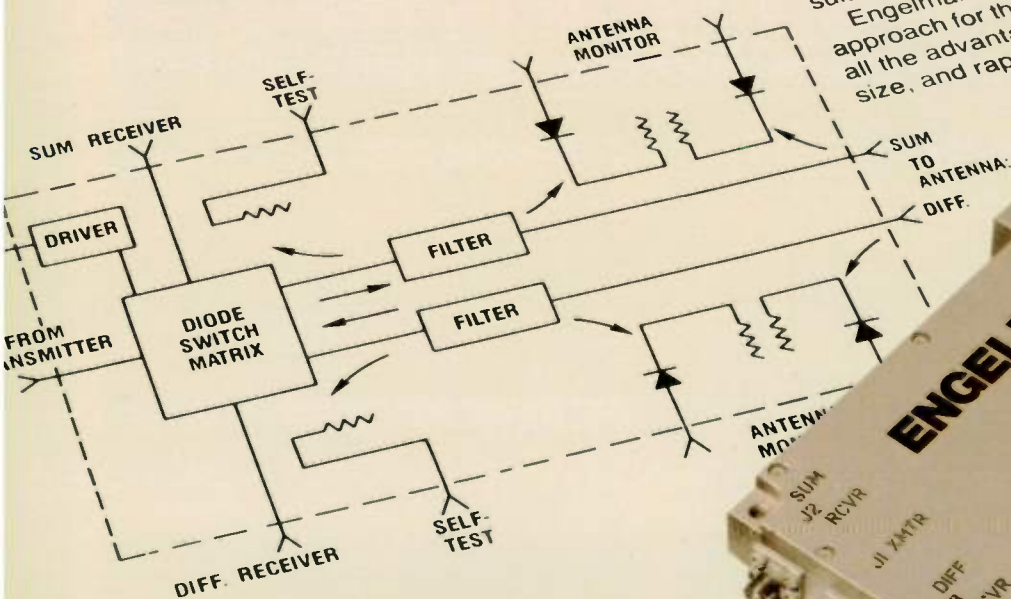
ISLS Switch and dual channel duplexing, filtering, antenna monitor detectors, and self-test coupler functions are all combined in this stripline RF module, with a total insertion loss of only 1.0db from antenna to either receiver.

The pin diode switch matrix connects 3.2 KW peak, 20 Watts average at 1030MHz transmitted power to two antennas in the aircraft system. The I.L. in this mode is 1.25db, and the switching speed is 200 nanosecs max. Low and band pass filters are provided for unwanted signal rejection out to 12GHz.

Forward and reflected directional detectors are provided to constantly detect and monitor transmitter and antenna efficiency in the system. In the receive mode, 1090MHz is duplexed to the sum and difference ports of the receiver.

Directional couplers are included in the receiver circuit to make a built-in test (BIT) circuit which allows the system operator to perform an operational check in any mode.

The design features a rugged modularized one-piece housing construction which is highly suitable for MIL-E-5400 Class 2 operation. Engelmann Microwave selected a stripline approach for this custom MIC because it offered all the advantages of low insertion loss, small size, and rapid and cost effective development.



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Dynamic Range

Model Number	Freq. Range (MHz)	Min. Gain (dB)	Gain Flat. (\pm dB)	Noise Figure		VSWR Max.		1 dB Gain Comp. (dBm Min.)	3rd Order Inter. (Typ.)	DC Power	
				Typ.	Max.	In	Out			Volt	mA
To 500 MHz											
AU-2A-0110	1-100	30	.5	1.0	1.3	2:1	2:1	0	+10	15	30
AU-3A-0110	1-100	45	.5	1.0	1.3	2:1	2:1	+5	+15	15	50
AU-4A-0110	1-100	60	.5	1.0	1.3	2:1	2:1	+10	+20	15	75
AU-2A-0120	1-200	30	.5	1.2	1.4	2:1	2:1	+5	+15	15	55
AU-3A-0120	1-200	45	.5	1.2	1.4	2:1	2:1	+5	+15	15	50
AU-4A-0120	1-200	60	.5	1.2	1.4	2:1	2:1	+10	+20	15	75
AU-2A-1045	100-450	30	.5	1.3	1.6	1.5:1	1.5:1	+7	+17	15	50
AU-3A-1045	100-450	45	.5	1.3	1.6	1.5:1	1.5:1	+10	+20	15	50
AU-4A-1045	100-450	60	.5	1.3	1.6	1.5:1	1.5:1	+10	+20	15	80
AU-1A-0150	1-500	15	.5	3.5	4.0	2:1	2:1	+10	+20	15	40
AU-2A-0150	1-500	30	.5	1.5	1.8	2:1	2:1	+10	+20	15	55
AU-3A-0150	1-500	45	.5	1.5	1.8	2:1	2:1	+10	+20	15	70
AU-4A-0150	1-500	60	.5	1.5	1.8	2:1	2:1	+10	+20	15	100
To 1000 MHz											
AM-1A-0510	500-1000	8	.5	2.0	2.5	2:1	2:1	-5	+5	15	20
AM-2A-0510	500-1000	20	.5	1.5	1.8	2:1	2:1	0	+5	15	45
AM-3A-0510	500-1000	30	.5	1.5	1.8	2:1	2:1	+5	+15	15	75
AM-4A-0510	500-1000	40	.5	1.5	1.8	2:1	2:1	+10	+20	15	105
AM-1A-000110	1-1000	10	.5	2.0	2.5	2:1	2:1	-10	0	15	20
AM-2A-000110	1-1000	25	.5	1.8	2.2	2:1	2:1	+5	+15	15	70
AM-3A-000110	1-1000	35	.5	1.8	2.2	2:1	2:1	+10	+20	15	60
AM-4A-000110	1-1000	50	.5	1.8	2.2	2:1	2:1	+10	+20	15	105



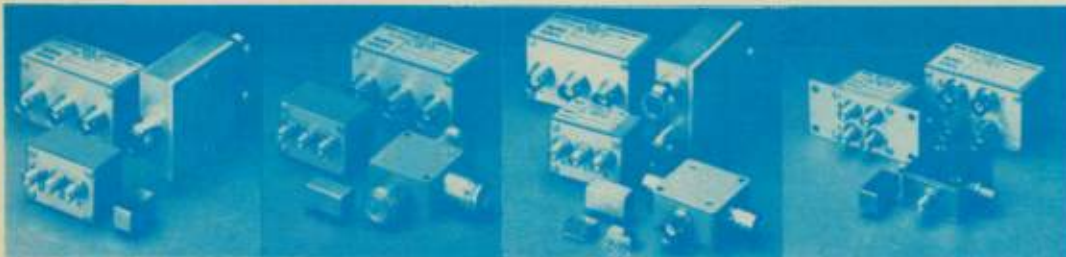
MITEQ AMPLIFIERS

To 1500 MHz											
AM-1A-0515	500-1500	10	.5	2.0	2.5	2:1	2:1	-5	+5	15	25
AM-2A-0515	500-1500	18	.5	2.0	2.5	2:1	2:1	-5	+5	15	50
AM-3A-0515	500-1500	30	.5	2.0	2.3	2:1	2:1	+7	+17	15	75
AM-4A-0515	500-1500	40	.5	2.0	2.3	2:1	2:1	+7	+17	15	100
AM-1A-000515	5-1500	10	.5	3.0	3.2	2:1	2:1	-10	0	15	20
AM-2A-000515	5-1500	25	.5	3.0	3.2	2:1	2:1	+10	+20	15	60
AM-3A-000515	5-1500	30	.5	3.0	3.2	2:1	2:1	+10	+20	15	100
AM-4A-000515	5-1500	40	.5	3.0	3.3	2:1	2:1	+10	+20	15	100
To 2000 MHz											
AM-1A-1020	1000-2000	10	.5	2.0	2.5	2:1	2:1	-10	0	15	10
AM-2A-1020	1000-2000	18	.5	2.0	2.5	2:1	2:1	0	+10	15	50
AM-3A-1020	1000-2000	30	.5	2.0	2.5	2:1	2:1	+10	+20	15	75
AM-4A-1020	1000-2000	35	.75	2.0	2.5	2:1	2:1	+10	+20	15	105
AM-5A-1020	1000-2000	50	.75	2.0	2.5	2:1	2:1	+10	+20	15	120
AM-3A-0322	300-2200	30	.75	2.2	2.75	2:1	2:1	0	+10	15	75
AM-1A-0420	400-2000	7	.5	2.4	2.8	2:1	2:1	-10	0	15	20
AM-2A-0420	400-2000	18	.5	2.0	2.5	2:1	2:1	0	+10	15	50
AM-3A-0420	400-2000	30	.5	2.0	2.5	2:1	2:1	+5	+15	15	75
AM-4A-0420	400-2000	40	.5	2.0	2.5	2:1	2:1	+5	+15	15	100
AM-5A-0420	400-2000	50	.75	2.0	2.5	2:1	2:1	+10	+20	15	125
AM-1A-000520	5-2000	7	.5	3.0	3.5	2:1	2:1	-10	0	15	20
AM-2A-000520	5-2000	15	.5	3.5	4.0	2:1	2:1	-10	0	15	50
AM-3A-000520	5-2000	23	.5	3.5	4.0	2:1	2:1	+5	+15	15	75
AM-4A-000520	5-2000	30	.75	3.5	4.2	2:1	2:1	+5	+15	15	100

Contact factory for custom options - often at no extra charge.

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TWO WAY 180°
(10 KHz-500 MHz)

TWO WAY 0°
(2 KHz-4200 MHz)

THREE WAY 0°
(0.01-750 MHz)

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For complete specifications and performance curves refer to the Microwaves Product Data Directory, EEM, or the Gold Book.

Model	Freq. range (MHz)	Min. isol.-dB (Mid-band)	Max. insert. loss.-dB (Mid-band)	See notes below	Price (Qty.)
2-WAY 90°					
PSCQ2 1.5	1.4-1.7	25	0.7 ¹	2	\$12.95 (5.49)
PSCQ2 3.4	3.0-3.8	25	0.7 ¹	2	\$16.95 (5.49)
PSCQ2 6.4	5.8-7.0	25	0.7 ¹	2	\$12.95 (5.49)
PSCQ2 7.5	7.0-8.0	25	0.7 ¹	2	\$12.95 (5.49)
PSCQ2 10.5	9.0-11.0	20	0.7 ¹	2	\$12.95 (5.49)
PSCQ2 13	12-14	25	0.7 ¹	2	\$12.95 (5.49)
PSCQ2 14	12-16	25	0.7 ¹	2	\$16.95 (5.49)
PSCQ2 21.4	20-23	25	0.7 ¹	2	\$12.95 (5.49)
PSCQ2 50	25-50	20	0.7 ¹	2	\$19.95 (5.49)
PSCQ 2.70	40-70	20	0.7 ¹	2	\$19.95 (5.49)
PSCQ 2.90	55-90	20	0.7 ¹	2	\$19.95 (5.49)
PSCQ 2.120	80-120	18	0.7 ¹	2	\$19.95 (5.49)
PSCQ 2.180	120-180	15	0.7 ¹	2	\$19.95 (5.49)
PSCQ 2.250	150-250	18	0.8 ²	2	\$19.95 (5.49)
PSCQ 2.400	250-400	16	0.9 ²	2	\$19.95 (5.49)
PSCQ 2.450	350-450	16	0.9 ²	2	\$19.95 (5.49)
ZSCQ 2.50	25-50	20	0.7 ¹	2,3	\$39.95 (4.24)
ZSCQ 2.90	55-90	20	0.7 ¹	2,3	\$39.95 (4.24)
ZSCQ 2.180	120-180	15	0.7 ¹	2,3	\$39.95 (4.24)
ZMSCQ 2.50	25-50	20	0.7 ¹	2,4	\$49.95 (4.24)
ZMSCQ 2.90	55-90	20	0.7 ¹	2,4	\$49.95 (4.24)
ZMSCQ 2.180	120-180	15	0.7 ¹	2,4	\$49.95 (4.24)
ZAPDQ 1	500-1000	20	0.9	2,13	\$59.95 (1.9)
ZAPDQ 2	1000-2000	18	0.9	2,13	\$59.95 (1.9)
ZAPDQ 4	2000-4200	20	0.9	2,13	\$59.95 (1.9)
2-WAY 180°					
PSCJ 2.1	1.200	25	0.8		\$19.95 (5.49)
PSCJ 2.2	0.01-20	25	0.5		\$29.95 (5.49)
ZSCJ 2.1	1.200	25	0.8	3	\$37.95 (4.24)
ZSCJ 2.2	0.01-20	25	0.5	3	\$47.95 (4.24)
ZMSCJ 2.1	1.200	25	0.8	4	\$47.95 (4.24)
ZMSCJ 2.2	0.01-20	25	0.5	4	\$57.95 (4.24)
ZFSCJ 2.1	1.500	25	1.5	5	\$49.95 (4.24)
ZFSCJ 2.3	5-300	25	1.5	5	\$39.95 (4.24)



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and combiners



FOUR WAY 0°
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SIX WAY 0°
(1-175 MHz)

EIGHT WAY 0°
(0.01-750 MHz)

SIXTEEN WAY 0° TWENTY FOUR WAY 0°
(0.5-125 MHz) (0.2-100 MHz)

Model	Freq. range (MHz)	Min. isol.-dB (Mid-band)	Max. insert. loss.-dB (Mid-band)	See notes below	Price (Qty.)	Model	Freq. range (MHz)	Min. isol.-dB (Mid-band)	Max. insert. loss.-dB (Mid-band)	See notes below	Price (Qty.)
2-WAY 0°						4-WAY 0°					
PSC-2-1	0.1-400	20	0.75		\$9.95 (6-49)	PSC-4-1	0.1-200	20	0.75		\$28.95 (6-49)
PSC-2-1W	1-650	20	0.9		\$14.95 (6-49)	PSC-4-1.75	1-200	20	0.9	1	\$24.95 (6-49)
PSC-2-2	0.002-60	20	0.6		\$19.95 (6-49)	PSC-4-3	0.25-250	20	0.75		\$23.95 (6-49)
PSC-2-1.75	0.25-300	20	0.75	1	\$11.95 (6-49)	PSC-4A-4	10-1000	15	1.1		\$49.95 (6-49)
PSC-2375	55-85	25	0.5	1	\$19.95 (6-24)	ZSC-4-1	0.01-40	25	0.5		\$29.95 (6-49)
PSC-2-4	10-1000	20	1.2		\$17.95 (6-49)	ZSC-4-1.75	0.1-200	20	0.75	3	\$46.95 (4-24)
MSC-2-1	0.1-450	20	0.75		\$16.95 (5-24)	ZSC-4-2	1-200	20	0.8	1.3	\$46.95 (4-24)
MSC-2-1W	2-650	25	0.8		\$17.95 (5-24)	ZSC-4-3	0.002-20	25	0.5	3	\$69.95 (4-24)
ZSC-2-1	0.1-400	20	0.75	3	\$27.95 (4-24)	ZMSC-4-1	0.25-250	20	0.75	3	\$43.95 (4-24)
ZSC-2-1.75	0.25-300	20	0.75	1.3	\$29.95 (4-24)	ZMSC-4-2	0.1-200	20	0.75	4	\$56.95 (4-24)
ZSC-2-1W	1-650	20	0.8	3	\$32.95 (4-24)	ZMSC-4-3	0.002-20	25	0.5	4	\$79.95 (4-24)
ZSC-2-2	0.002-60	20	0.6	3	\$37.95 (4-24)	ZFSC-4-1	0.25-250	20	0.75	4	\$53.95 (4-24)
ZSC-2375	55-85	25	0.5	1.3	\$37.95 (4-24)	ZFSC-4-1W	1-1000	18	1.5	8	\$89.95 (1-4)
ZMSC-2-1	0.1-400	20	0.75	4	\$37.95 (4-24)	ZFSC-4375	10-500	20	1.5	8	\$74.95 (1-4)
ZMSC-2-1W	1-650	20	0.8	4	\$42.95 (4-24)	ZA4PD-2	50-90	30	1.2	1.8	\$89.95 (1-4)
ZMSC-2-2	0.002-60	20	0.6	4	\$47.95 (4-24)	ZA4PD-4	1000-2000	18	1.0	14	\$79.95 (1-9)
ZFSC-2-1	5-500	20	0.6	5	\$31.95 (4-24)		2000-4200	18	1.0	14	\$79.95 (1-9)
ZFSC-2-1.75	0.25-300	20	0.75	5	\$32.95 (4-24)	6-WAY 0°					
ZFSC-2-1W	1-750	20	0.8	5	\$35.95 (4-24)	PSC-6-1	1-175	18	1.0		\$68.95 (1-5)
ZFSC-2-2	10-1000	20	1.0	5	\$39.95 (4-24)	ZFSC-6-1	1-175	20	1.2	9	\$89.95 (1-4)
ZFSC-2-4	0.2-1000	20	1.0	5	\$44.95 (4-24)	8-WAY 0°					
ZFSC-2-5	10-1500	20	1.0	5	\$49.95 (4-24)	PSC-8-1	0.5-175	20	1.1		\$68.95 (1-5)
ZFSC-2-6	0.002-60	20	0.6	5	\$36.95 (4-24)	PSC-8-1.75	0.5-175	20	0.8	1	\$69.95 (1-5)
ZFSC-2-6.75	0.004-60	20	0.8	1.5	\$38.95 (4-24)	PSC-8A-4	5-500	18	1.8		\$89.95 (1-5)
ZAPD-1	500-1000	19	0.6	6	\$39.95 (1-9)	PSC-8-6	0.01-10	23	1.1		\$79.95 (1-5)
ZAPD-2	1000-2000	19	0.6	6	\$39.95 (1-9)	ZFSC-8-1	0.5-175	20	1.1	10	\$89.95 (1-4)
ZAPD-21	500-2000	18	0.7	6	\$49.95 (1-9)	ZFSC-8-1.75	0.5-175	20	1.0	1.10	\$90.95 (1-4)
ZAPD-4	2000-4200	19	0.8	6	\$39.95 (1-9)	ZFSC-8375	50-90	25	1.3	1.10	\$119.95 (1-4)
3-WAY 0°						16-WAY 0°					
PSC-3-1	1-200	25	0.7		\$19.95 (5-49)	ZFSC-8-4	0.5-700	20	1.5	10	\$129.95 (1-4)
PSC-3-1W	5-500	15	1.4		\$29.95 (5-49)	ZFSC-8-6	0.01-10	23	1.1	10	\$109.95 (1-4)
PSC-3-1.75	1-200	25	0.7	1	\$20.95 (5-49)	24-WAY 0°					
PSC-3-2	0.01-30	25	0.45		\$29.95 (5-49)	ZFSC-16-1	0.5-125	18	1.6	11	\$174.95 (1-4)
PSC-3-13	1-200	35	0.6		\$24.95 (5-49)	24-WAY 0°					
ZSC-3-1	1-200	25	0.7	3	\$37.95 (4-24)	ZFSC-24-1	0.2-100	20	2.0	12	\$264.95 (1-4)
ZSC-3-1.75	1-200	25	0.7	1.3	\$38.95 (4-24)						
ZSC-3-2	0.01-30	25	0.45	3	\$47.95 (4-24)						
ZSC-3-2.75	0.02-20	25	0.6	1.3	\$48.95 (4-24)						
ZMSC-3-1	1-200	25	0.7	4	\$47.95 (4-24)						
ZMSC-3-2	0.01-30	25	0.45	4	\$57.95 (4-24)						
ZFSC-3-1	1-500	20	0.9	5	\$39.95 (4-24)						
ZFSC-3-1W	2-750	20	1.0	5	\$41.95 (4-24)						
ZFSC-3-13	1-200	35	0.6	5	\$39.95 (4-24)						

7 TNC, SMA & Type N at \$5 additional cost. Please specify connectors.
8 SMA connectors standard. BNC on request.
9 BNC connectors standard. TNC available. SMA available at \$15 additional cost.

10 BNC connectors standard. TNC available at \$10 additional cost. SMA at \$25 additional cost.
11 BNC connectors standard. TNC available at \$20 additional cost. SMA available at \$45 additional cost.

12 BNC connectors standard. TNC available at \$35 additional cost. SMA available at \$65 additional cost.
13 BNC connectors (not available for ZAPDQ-4). TNC available. (SMA [3MM] and Type "N" on request. Add \$5 per unit.) Please specify connectors.

14 TNC, SMA, Type N please specify connectors

Amplifiers

- Ultra Low Noise
- Wide Dynamic Range
- Custom Design

Model Number	Frequency (MHz)	Min. Gain (dB)	Flatness (dB)	Noise Figure (dB)		Pwr. Out @ 1 dB Compression Pt. (dBm)	Case/Connectors*
				typ.	max.		
W50ETD	0.01-50	50	±.5	1.3	1.5	0	C/SMA
W50ETC	0.01-50	20	±.5	4.0	4.5	+23	C/SMA
W250G	5-250	43	±.5	1.3	1.5	+25	B/SMA
W500E	5-500	30	±.5	1.3	1.4	0	C/SMA
L60E-2	50-70	60	±.5	0.9	1.0	+10	C/SMA
L450E	400-500	27	±.5	1.2	1.4	+5	C/SMA
W1G2H	5-1000	30	±.5	1.3	1.5	+5	C/SMA
W2GHH2	1-2 GHz	30	±.5	2.3	2.5	+5	AB/SMA

Ultra Low Noise Amplifiers

Special Purpose Amplifiers

Model Number	Frequency (GHz)	Gain (dB)	Noise Figure (dB)	Pwr. Out @ 1 dB Compression Pt. (dBm)	Case/Connectors
L13GE	1.25-1.35	25	2.2	+5	C/SMA
W89DGA	0.47-0.89	25	2.0	+5	C/SMA
L215GA	2.15-2.165	11	3.2	-3	C/N
L215GC	2.15-2.165	29	2.9	+7	C/N
W2GH	0.5-2.0	25	3.0	+10	B/SMA
P150P	0.08-150 MHz	60	1.5	+30	H/BNC
W15GB1	0.05-1.5	20	1.8	-3	C/SMA
W23GA	0.1-2.3	8	9.0	+20	C/SMA

Model Number	Frequency (GHz)	Min. Gain (dB)	Pwr. Out @ 1 dB Compression Pt. (dBm)		Noise Figure (dB)	Case/Connectors	Typical Intercept Pt. (dBm)
			typ.	min.			
P60F	30-90 MHz	30	+32	+31	5.5	H/BNC	+43
P150H2	0.1-150 MHz	27	+31.5	+30	6.5	H/BNC	+44
P175M	150-200 MHz	23	+34	+33	8.0	H/BNC	+45
P400C	10-400 MHz	20	+31	+30	7.0	H/BNC	+42
P500N	2-500 MHz	17	+31	+30	8.0	H/BNC	+42
P10GL	0.5-1.0	30	+31	+30	5.0	H/SMA	+42
P2GS-7	0.5-2.0 GHz	30	+30	+29	10.0	FS/SMA	+42
P24GB	1.4-2.4	16	+20	+19	8.0	A/SMA	+32

Wide Dynamic Range Amplifiers

CASE DIMENSIONS:
(Others Available)

	L (in.)	W (in.)	H (in.)
C	1.875	1.875	0.465
A	3.375	1.875	0.465
H	3.75	2.60	1.95
AB	3.00	1.875	0.465
B	2.625	1.875	0.465
FS	4.5	2.8	1.1

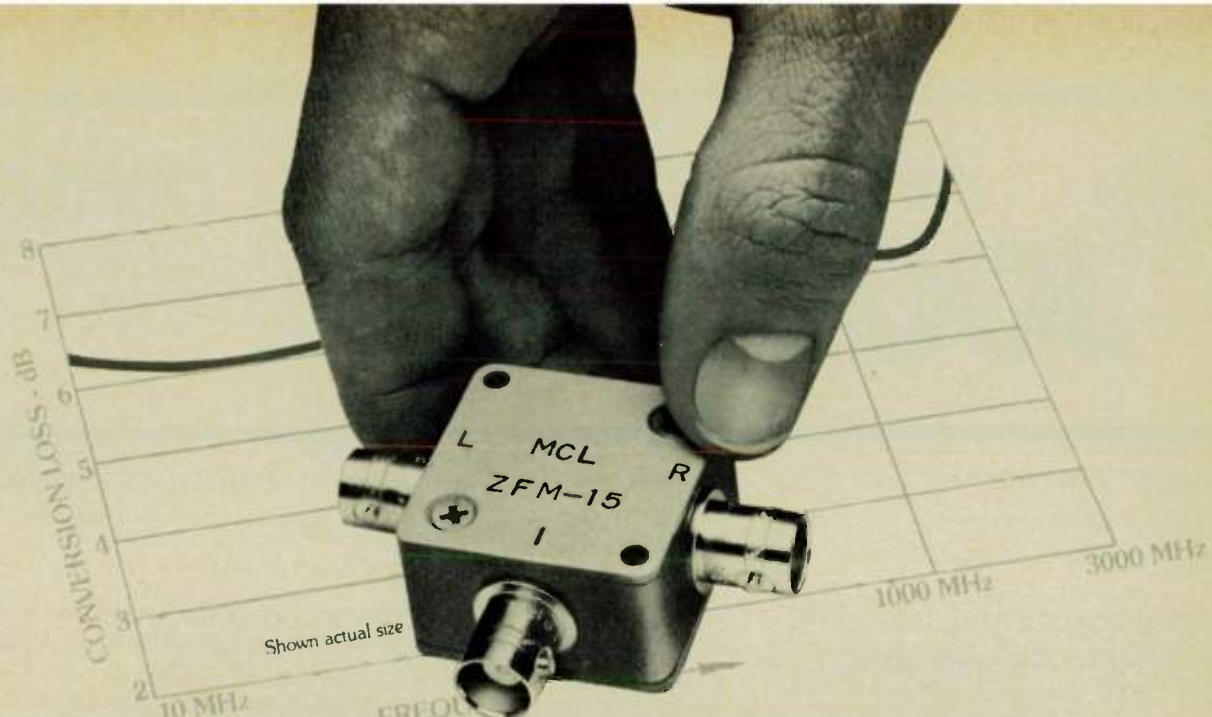
*Standard this model; others may be specified.
VSWR all models:
2:1 max, 1.5:1 typ.

The devices above are just a sampling of TRONTECH's product line in state of the art low noise and medium power amplifiers. Our amplifiers are designed so that parameters such as gain, bandwidth, output power, noise figure, form factor, etc., can be tailored to your specifications. Our capabilities include Video and Limiting Amplifiers, Integrated Assemblies, Modular Amplifiers, etc.



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ZFM 15 SPECIFICATIONS

FREQUENCY RANGE, MHz
 LO 10-3000 RF 10-3000 IF 10-800
 LO Power + 10 dBm

CONVERSION LOSS, dB	Typ	Max
20-1500 MHz	6.3	7.5
10-3000 MHz	6.5	8.5

ISOLATION, dB (10-3000 MHz)	Typ	Min
LO RF	30	20
LO IF	30	20

SIGNAL 1 dB Compression Level +5 dBm
 IMPEDANCE all ports 50 ohms

VSWR (at ports)	Typ
LO	2.1
RF	2.1
IF	1.8

CONNECTORS BNC STD. TNC on request
 Type N and SMA \$5 additional

MOUNTING Thru hole. Threaded insert. Flange
 SIZE 125 x 125 sq x 0.75" high

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CIRCLE 7 ON READER SERVICE CARD



microwave JOURNAL

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MAY 1981

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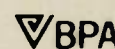
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World Radio History

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lightweight 95 GHz magnetron

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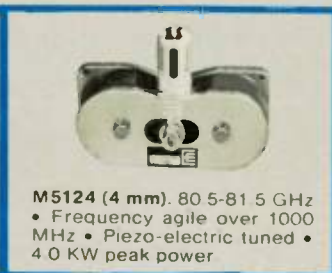


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94-96 GHz**
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only 4.0 lbs.

Actual Size



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World Radio History

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Coming Events

**1981 IEEE/MTT-S
INT'L MICROWAVE
SYMPOSIUM**
JUNE 15-17, 1981

Sponsor: IEEE
MTT-S (held jointly
with IEEE AP-S
and URSI on June
17-19, 1981).

Place: Bonaventure Hotel, Los Angeles, CA.
Theme: "Around the World with Micro-
waves," includes such topics as CAD and
measurement techniques, microwave and
mm-wave solid-state devices, IC's, low noise
techniques, mw passive components and
networks, microwave ferrite devices, satellite
communications, submm-wave techniques
and devices, mw bioeffects, etc. Contact:
Al Clavin, Hughes Aircraft Co., Bldg. 268/
A-55, Canoga Park, CA 91304. Tel: (213)
702-1778.

**1981 IEEE INT'L
SYMPOSIUM ON
EM COMPATIBILITY**
AUG. 18-20, 1981

Sponsor: IEEE
Place: Boulder, CO
Subject: Problems
of electromagnetic
compatibility and

electromagnetic interference, including EM
properties of composites, FCC measure-
ments, EM compatibility design in micro-
electronics, etc. Contact: Charlotte Tyson,
Registration EMC '81, Int'l Business Ma-
chines, 59Z/025-1, P.O. Box 1900, Boulder,
CO. 80302. Tel: (303) 447-5072.

**1981 ANTENNA
MEASUREMENT
TECHNIQUES
ASSN. SYMPOSIUM**
OCT. 13-15, 1981

Call for Papers.
Sponsor: Antenna
Measurement Tech-
niques Association,
host: Sanders Assoc.
Place: King's Grant

inn, Danvers, MA. Topics: Near Field Test-
ing, Compact Antenna Ranges, Advanced
Instrumentation and Automation of Antenna
Measurements. Submit four copies of a
brief abstract by June 15, 1981 to George
Andrews, AMTA Sec., Boeing Aerospace
Co., P.O. Box 3999, M/S 8E-17, Seattle, WA
98124. Tel: (206) 773-9811. Gen. Contact:
Ivan Davis, Chrmn., Scientific-Atlanta, Inc.,
Instrumentation Div., Atlanta, GA 30340.
Tel: (404) 449-2038.

**1981 IR AND
MM WAVE CONF.**
DEC. 7-12, 1981

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Microwave Theory
and Techniques

Society. Place: Carillon Hotel, Miami Beach,
FL. Sessions: including mm sources, de-
vices or system, mm and sub-mm propaga-
tion, spectroscopy, lasers, imaging, etc. Sub-
mit 35-40 word abstract by June 30, 1981
to: K. J. Button, M.I.T., National Magnet
Lab., Cambridge, MA 02139. Tel: (617)
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C-002	High Speed SPST switch modules with Internal Bias Circuitry
C-003	High Speed SPST Microwave Switch Modules
C-004	Medium Power SPST Switch Modules
C-005	Series Type SPMT Switch Modules
C-006	Series Type SPDT Switch Modules
C-007	Non Reflective SPMT Switch Modules
C-008	Series Type SPDT Modules with Internal Bias
C-009	Series Type Non Reflective SPDT Switch Modules w/Int. Bias
C-010	Single Diode Switch Modules
C-011	Ultra-Fast SPDT Switch Modules
C-012	Ultra-Fast SPMT Switch Modules
C-013	High Speed SPST Switches with Drivers
C-014	General Purpose SPST Switches

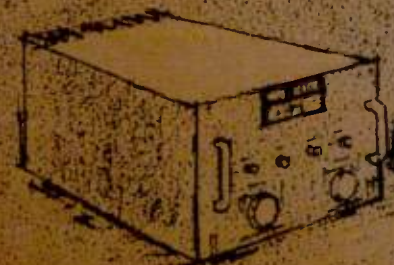
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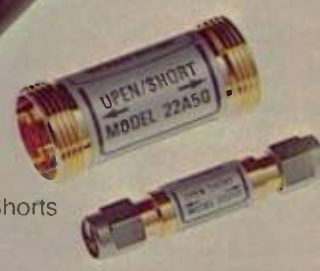
SWR Autotesters and SWR Bridges



Precision Adapters



Precision Air Lines



Open/Shorts

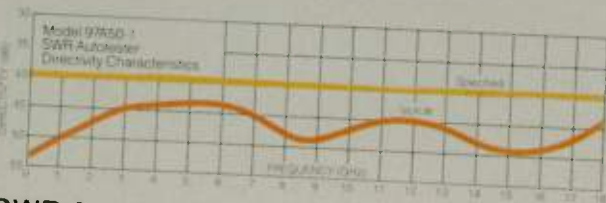


Precision Terminations

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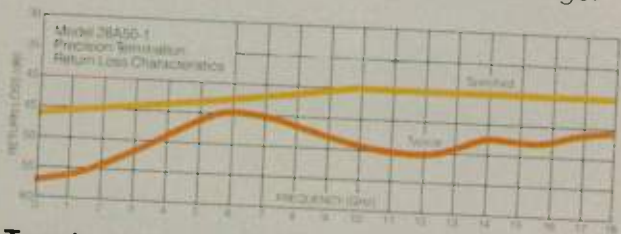
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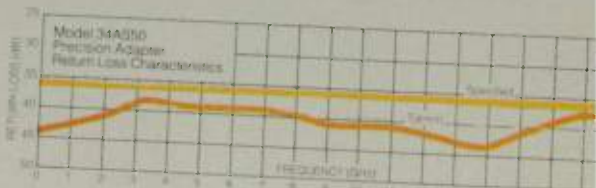


Terminations and Open/Shorts

Wiltron Terminations provide an accurate reference for SWR measurements as well as a termination for test instruments and devices under test from DC to 26.5 GHz. They are available in GPC-7, N and WSMA connectors and feature aged termination resistors for long-term stability. Maximum SWR varies from 1.002 at low frequencies to 1.135 at 26.5 GHz. Wiltron 22 Series Open/Shorts for the DC to 18 GHz range are offered with a choice of connectors.

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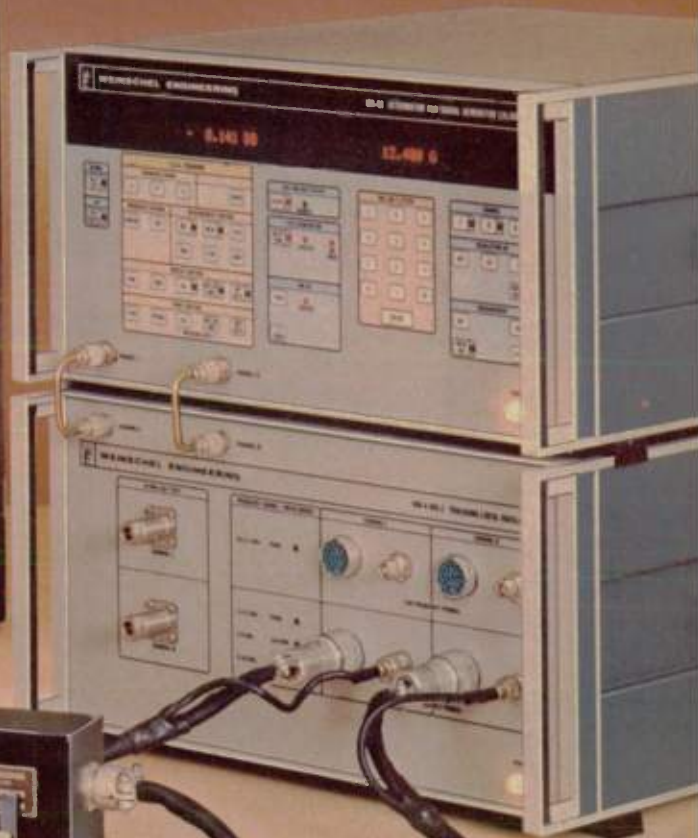
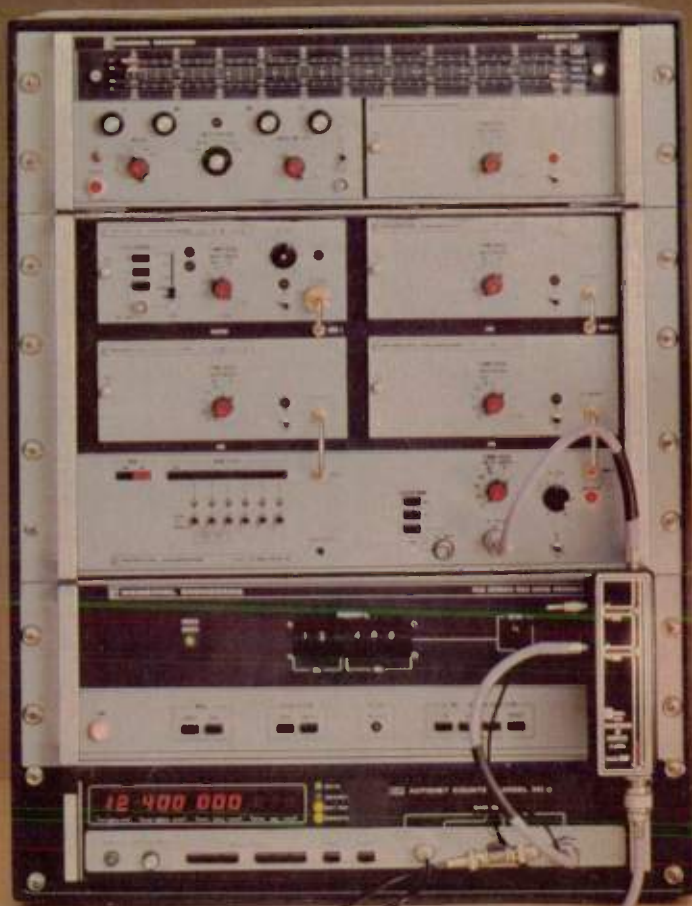
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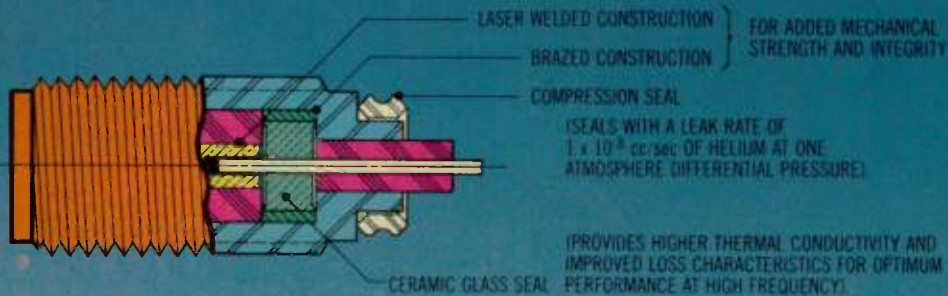


IEEE-488



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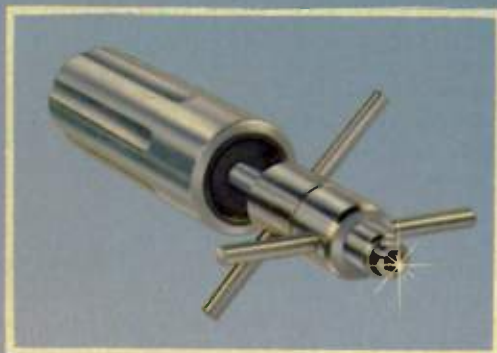
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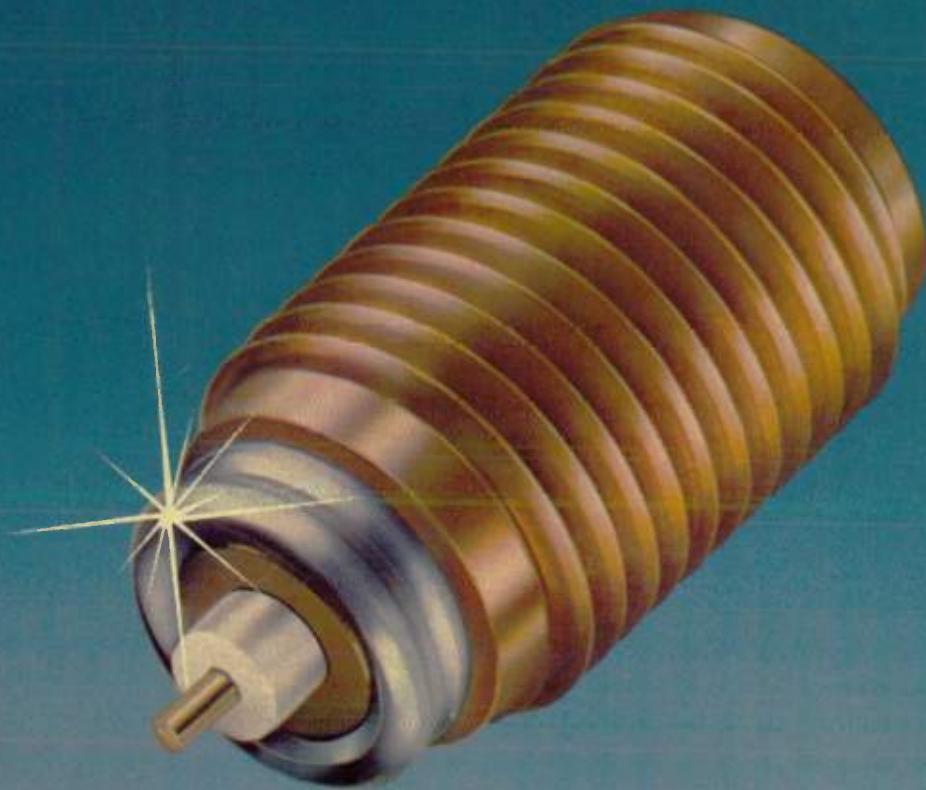


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- Thread connector into MIC package
- Tighten until torque wrench releases

...The OSCS Series



ACTUAL SIZE



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World Radio History



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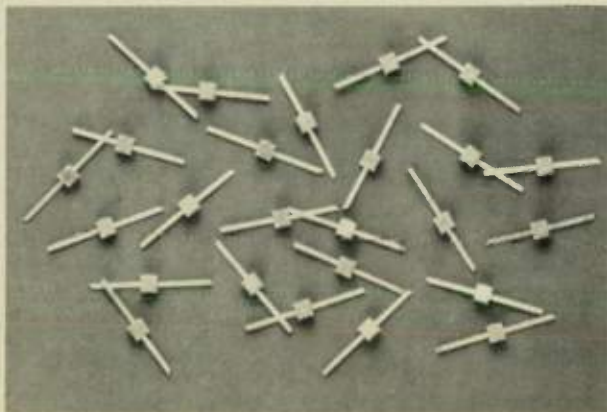


ELECTRICAL CHARACTERISTICS

Part Number	I _p (μa)	V _r (Mv) Typ.	V _f (Mv) Typ.	R _s (Ω) Max.	C _t (pf) Max.
23100BD	50-150	540	120	12	7
23150BD	50-150	540	100	9	1.0
23200BD	150-250	550	110	11	7
23250BD	150-250	550	100	8	1.0
23300BD	250-350	540	110	11	7
23350BD	250-350	540	100	8	1.0
23400BD	400-500	530	100	10	7
23450BD	400-500	530	80	8	1.0

ENVIRONMENTAL RATINGS

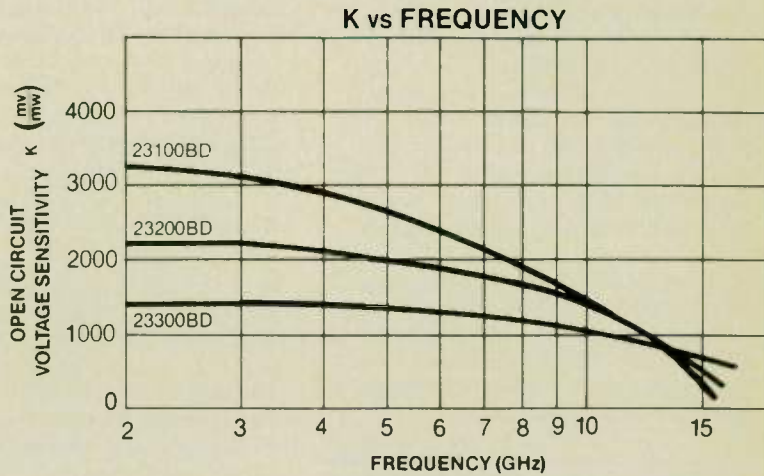
Storage Temp. —65°C to +100°C
 Operating Temp. —65°C to +100°C
 Temperature Cycle—MIL-STD-202, Method 107, TC. A
 Shock—MIL-STD-750, Method 2016
 Vibration—MIL-STD-750, Method 2056
 Acceleration—MIL-STD-750, Method 2056
 Fine Leak—MIL-STD-750, Method 1071, TC. G
 Gross Leak—MIL-STD-750, Method 1071, TC. C



Typical Performance

Type	R _v (Ω)	T _{SS} (-dbm)	M ($\sqrt{\frac{K}{R_v}}$)	K ($\frac{mv}{mw}$)
23100BD	400	-52	80	1600
23200BD	180	-51	90	1200
23300BD	80	-50	100	900

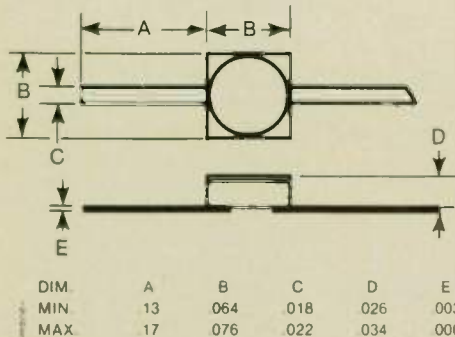
NOTES: 1. 2MHz Band Width 2. Input Power = 20 dbm 3. F_o 10GHz



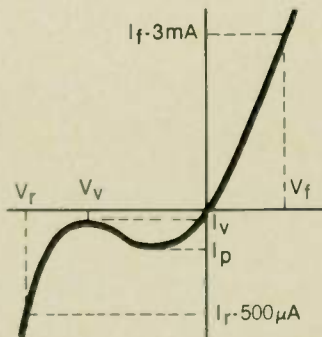
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PACKAGE OUTLINE Type 23



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1981 MICROWAVE MEETINGS

This year's MTT-S Symposium/Exhibition will be augmented by the coincident AP-S and URSI Symposia. All three meetings will be held during the week of June 15-19 at the Bonaventure Hotel in Los Angeles and, together with the Microwave Exhibition which will run from June 15 through 17, this promises to be the largest gathering ever of the microwave industry. Beginning with Joe White's "Attending the Conference" article, a major portion of this issue is devoted to meeting and exhibition information. Complete programs for the MTT-S, AP-S and URSI Symposia are listed. Information about evening panel sessions and special workshops is included. Complete floor plans of the meeting and exhibition and an exhibition guide are also provided.

Sum Up



I.E.E. MICROWAVE CIRCUIT COMPONENT MEETING

For the first time in approximately 10 years, the professional group on Microwave Devices and Techniques of I.E.E. convened a meeting in December in London to review microwave circuit components. The first of two sessions addressed GAs FET's for microwave signal control, optical control of microwave semiconductors and ferrite control devices. The second session concentrated on PIN diode applications for modulation, control and switching.

MARCHAND BALUN CIRCUIT ELEMENT GRAPHS

Implementing transformer action within a Marchand balun to match the balanced load resistance of a typical antenna to the source resistance leads to optimum broadband characteristics. Graphs of the circuit element values for a fourth order balun with single section quarter wave input and output transformers are provided in this article. Bandwidth ratios ranging from 4:1 to 12:1 are covered.

AN ARCSPRAYED COATING FOR EMI/RFI SHIELDING

A system for applying molten metal particles to a surface for EMI/RFI shielding purposes which does not appreciably raise the temperature of the treated surface is described. Attenuation characteristics of the sprayed material through 10 GHz are discussed. Mechanical properties of the material and a complete description of the equipment and its operation are also included.

THE FRENCH MICROWAVE LANDING SYSTEM

The history of the development of the microwave landing system is discussed and its several advantages over present systems are described. The transition plan from ILS to MLS as currently defined by the International Civil Aviation Organization which calls for general use of MLS during the 1995 to 2000 time frame is shown. The French system presently being manufactured by Thomson-CSF is covered in detail. The design approach and performance of major subsystems is described.

Howard Ellavitz

Workshops & Courses

MICROWAVE ANTENNA MEASUREMENTS SHORT COURSE

Sponsor: California State U., Northridge
Site: CSU, Northridge, CA
Date: July 13-17, 1981
Fee: \$575
Description: Measurements of microwave antenna characteristics, including planar, cylindrical and spherical near-field techniques.
Contact: Dr. Edmond S. Gillespie, School of Engineering and Computer Science CSU, Northridge 18111 Nordhoff Street Northridge, CA 91330 Tel: (213) 885-2190, 885-2146

MICROWAVE SYSTEMS PLANNING

Sponsor: George Washington U. Continuing Engrg. Ed.
Site: GWU, Washington, DC
Date: July 13-17, 1981
Fee: \$715- No. 832DC
Instructor: Allan Schneider, Pres. CyberCom Corp.
Description: Concepts and principles of microwave system planning, including the impact of digital technology on mw transmission systems.
Contact: Continuing Engineering Education, George Washington University Washington, DC 20052

GIT SHORT COURSES

Sponsor: Georgia Institute of Technology
Site: Georgia Tech., Atlanta, GA
Date: July 20-22, 1981
Fee: \$350
Subject: Millimeter and Microwave Ferrite Materials—Study of the structure and properties of ferrite materials and the application of these materials in microwave signal processing and control functions.
Contact: Department of Continuing Education, Georgia Institute of Technology, Atlanta, GA 30332 Tel: (404) 894-2400

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WJ-6855 Series	4 to 8 GHz	WJ-6859 Series	12 to 20 GHz



ATTENDING THE CONFERENCE



JOSEPH F. WHITE, Consulting Editor

THE BONAVENTURE SITE

Coming to this year's MTT Conference at the Bonaventure Hotel in Los Angeles will be like making your own space odyssey. The Bonaventure with 1500 guest rooms, half-acre exhibit space, 37 meeting rooms and 4 ball-rooms is a twenty-three story symmetric arrangement of five silo-like structures, a center one encircled by four others with a six-story skylight lobby enclosing registration, bars, restaurants, boutiques, and gardens. Adjacent to this artistic inn by architect John Portman, is a multi-story conference center with roof swimming pool. On the first floor of this exhibit structure will be found over 150 separate microwave exhibits and booths. On the next two floors are the meeting rooms in which up to four simultaneous sessions containing 160 presentations over three days will be held. The drawing on p. 30 is our attempt to help you orient yourself as you traverse the maize of corridors and towers that intervenes between your room and the papers you want to hear.

The Bonaventure is located as strategically in downtown LA as could be wished, which translated means you'll need a car, bus or taxi to get to most places of interest which aren't immediate to the hotel complex. (See L.A. Area Map, p. 28).

Conference registration (for any one or all — MTT, AP and URSI) will take place in the foyer of the Bonaventure in front of the California Ballroom from 4 PM to 9 PM on Sunday, from 7:30 AM to 9 PM on Monday, from 7:30 AM to 4 PM Tuesday through Thursday and from 7:30 AM to 12 PM on Friday. Messages can be transmitted to conference attendees through the hotel operators by calling the toll-free numbers USA 800-228-3000, Alaska and Hawaii 800-228-1212, Canada 800-268-8383 giving the name of the person to be contacted and indi-

cating that that person is an IEEE conference attendee. Those who cannot be accommodated in the Bonaventure Hotel will be housed at the Biltmore.

There are seven different establishments within the Bonaventure with restaurant facilities. Buffet breakfast and lunches are served in the lobby court and there is a variety of restaurants within a few walking blocks of the hotel.

SPECIAL PANEL SESSIONS

This year's evening panel talks highlight the wide diversity of questions besetting the microwave industry. Monday night features "Status and Trends in Fiber Communications," a session headed by M. K. Barnoski and J. Austin of the TRW Technology Research Center describing the background of fiber optic communications, its capabilities and applications, and its relationship to competing microwave links in the market place. Actual installations and systems will be presented with experts from both the fields of fiber optics as well as microwave communications to present aspects of fiber optic communications and to respond to audience.

Another Monday evening panel covers "Microwave Devices Technology Transfer from Research and Development to Manufacturing." This panel will address the production syndrome created within the microwave industry which results because microwave devices enjoy relatively little production volume yet must meet tighter performance specifications, more stringent material parameter tolerances and more complex fabrication and testing than other goods produced within an industrial nation. Problems and challenges for automation, labor training, capital investment and cost effectiveness will be discussed by the panel, headed by John Kuno of Hughes Aircraft Company.

The "Impact of Gallium Arsenide IC's on Microwave Systems" is the subject of a Panel (meeting Wednesday night 1) organized by Paul T. Greiling.

The panel members encourage audience participation in this discussion which will relate the performance of state of the art digital and analog GaAs IC's along with their potential for lower cost, lower noise, wider bandwidth or higher data rates and how these desirable performance features may be limited by material quality, power dissipation, on and off chip delays, and packaging.

Wednesday night, "Manpower Problems in Microwave and Antenna Engineering," will be addressed in a panel headed by R. McIntosh, NASA, Langley, Virginia.

Also on Wednesday night, "State of the Art of Millimeter and Submillimeter Receivers," a panel organized by J. J. Whelehan of AIL, Melville, NY will treat the evolving state of the millimeter and submillimeter technology and the need for system trade-off decision by potential user groups which exists at this developing state of the millimeter art. A special topic will include the overlap region between millimeter and submillimeter-wave receivers.

MTT TECHNICAL PROGRAM

The 1981 MTT Symposium Technical Program includes 22 separate sessions containing over 150 papers which will be presented in four days in up to 4 simultaneous technical sessions at any one time. The old standby subjects including Computer Aided Designs, Ferrites, Passive Components, Phased Array Techniques, Low Noise Techniques and Field and Network Theory account for about half the papers. Equal representation is given newer topics of interest to the microwave community including FET's and associated Solid State devices, Millimeter Waves, 6-Port Measurement Methods, Dielectric Resonators, Printed and Monolithic Integrated Circuits, Microwave Acoustics, Guided Wave Optics and Bio-Medical Effects and Applications.

The FET papers cover some impressive accomplishments. For example, Session B includes an FET with a 0.6 dB noise figure at 4 GHz, 1.3 dB

(continued on page 28)

Sand dunes at White Sands, New Mexico
No. 1 in a series of abstract waveforms
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Hughes is making new waves with a shortcut to mm-wave radar design.

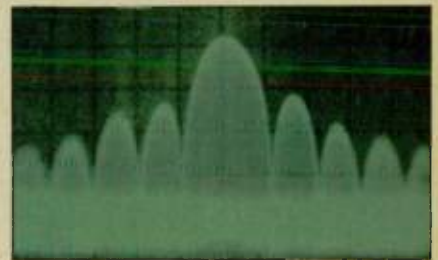
Now designers don't have to reinvent the radio when they take on millimeter-wave radar development. That's because Hughes short cuts the millimeter-wave learning curve by offering preassembled and tested subsystems right off the shelf.

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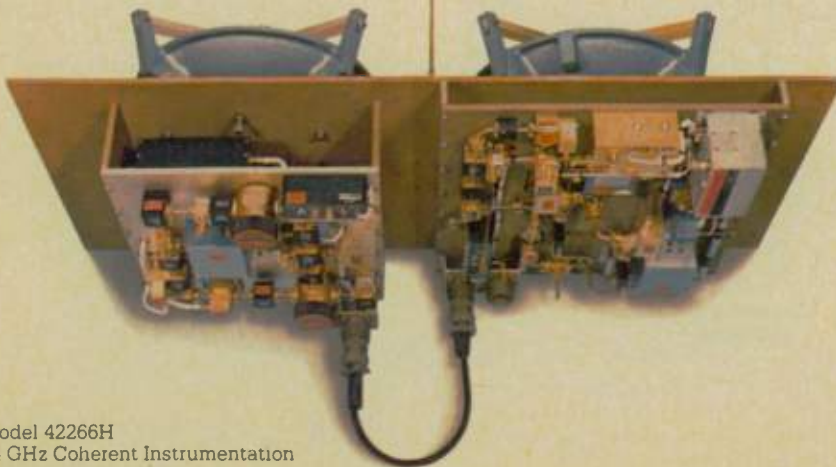
Operating at 94 GHz this fully calibrated breadboard combines transmitter, receiver and antenna functions all in one subsystem. It performs all mixing, detecting, switching and modulating functions needed to provide an S-band IF output signal.

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Model 42266H
94 GHz Coherent Instrumentation
Radar Front End

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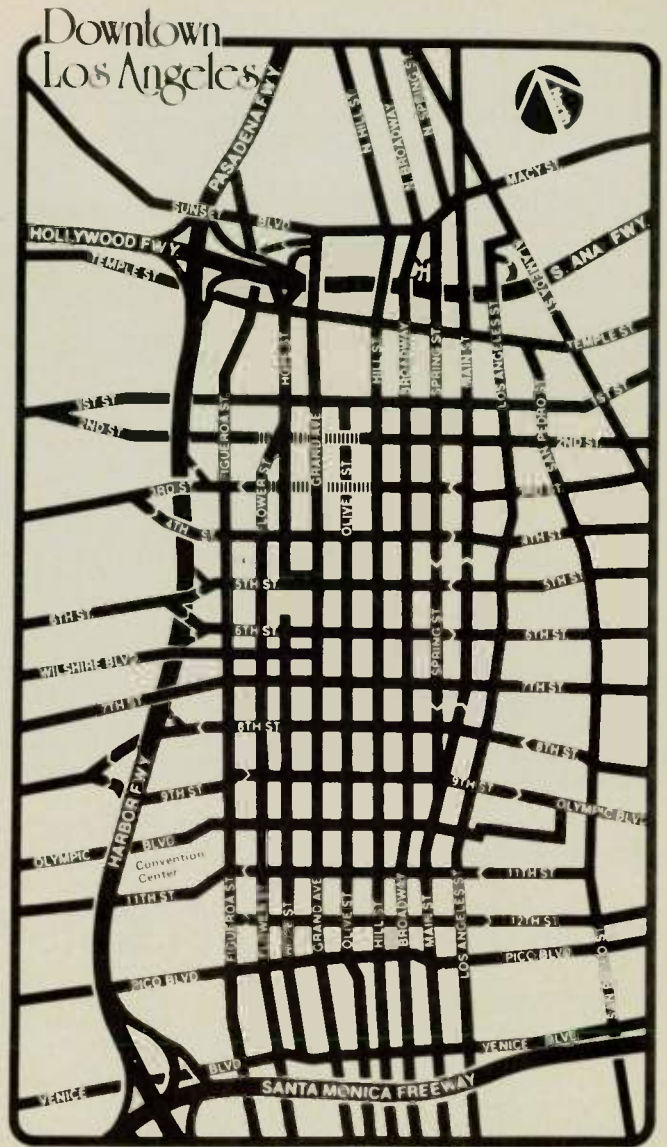
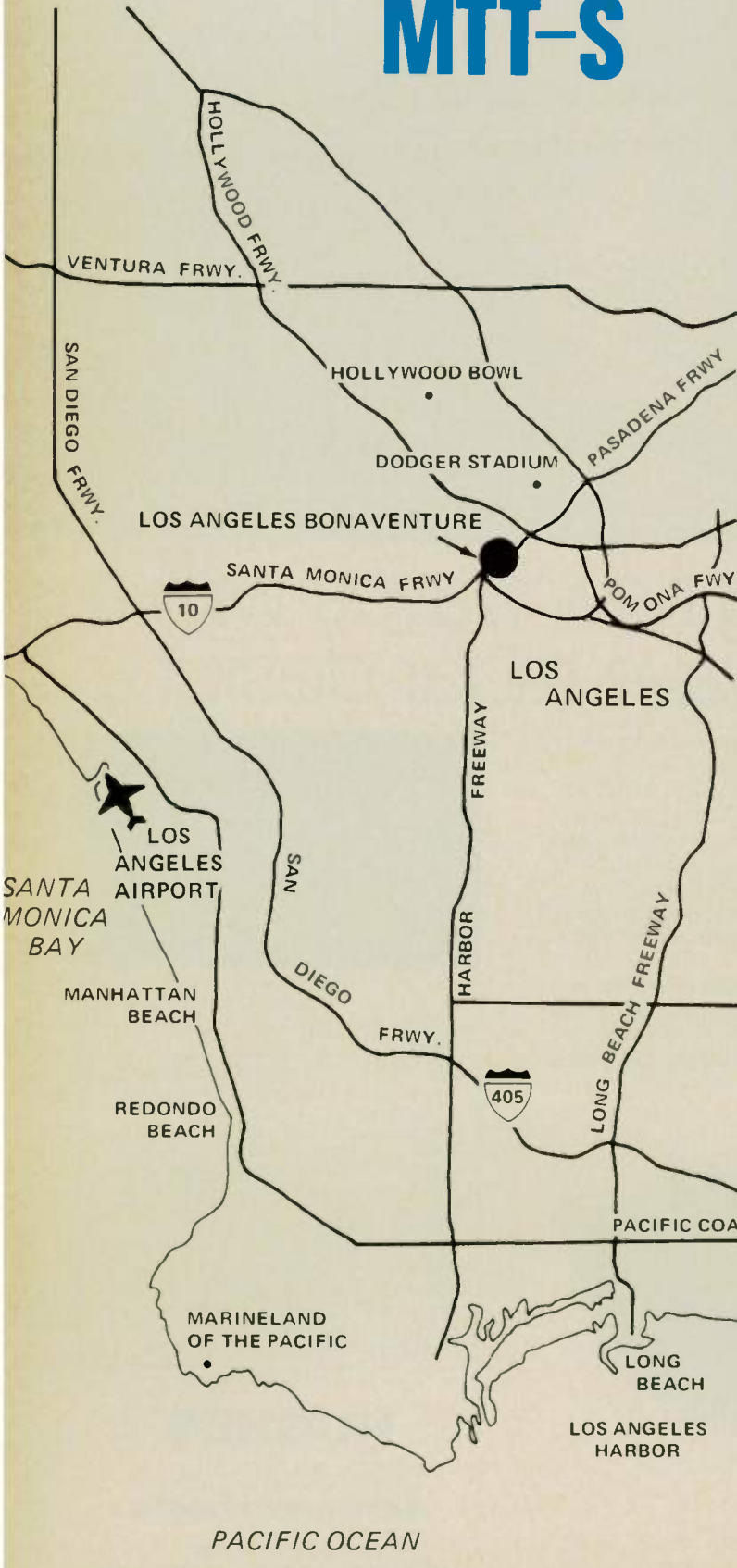
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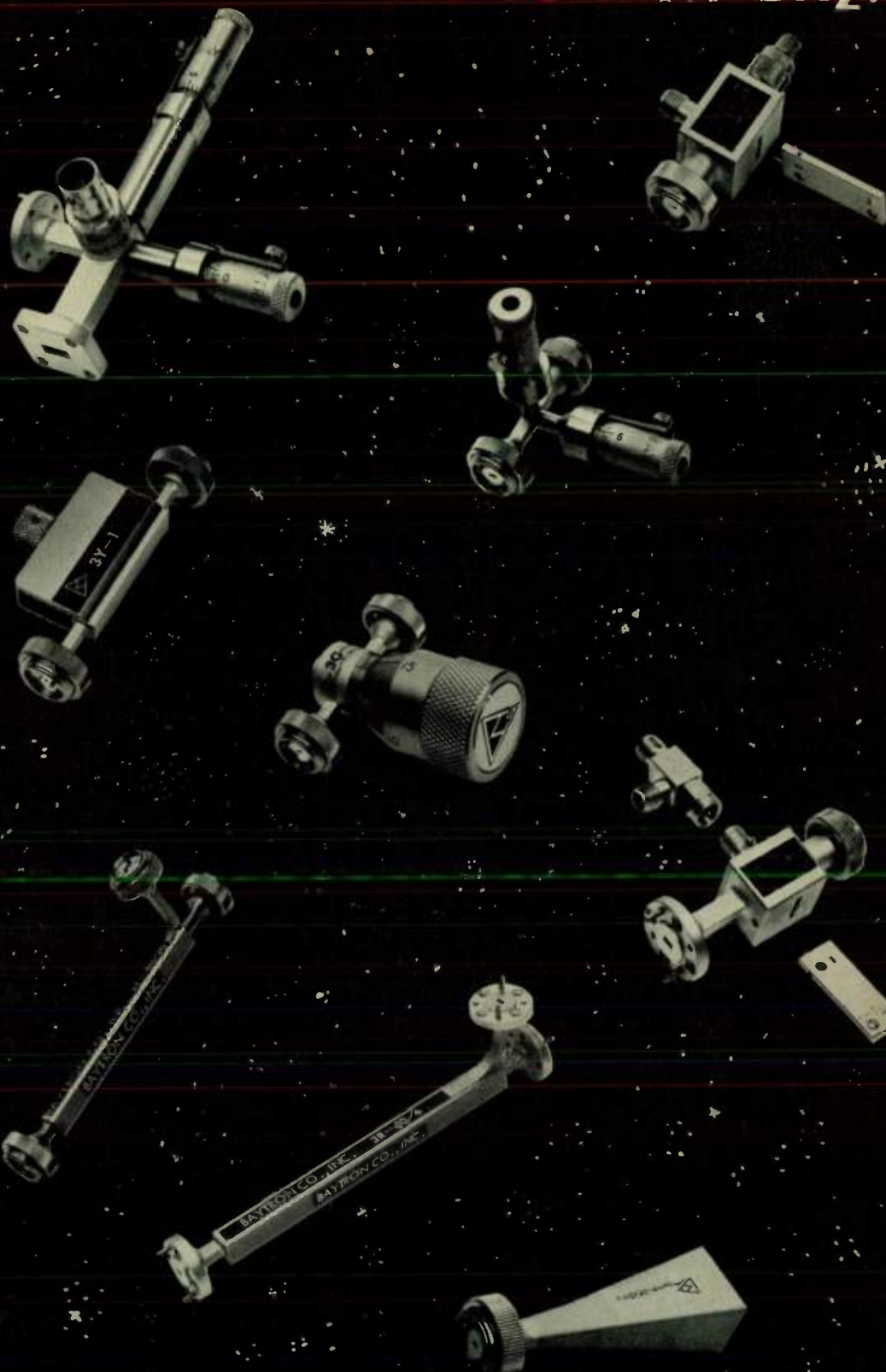
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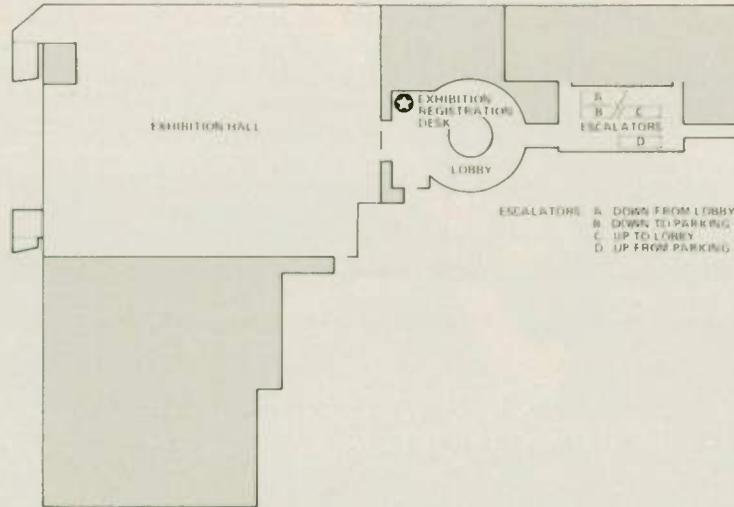
World Radio History

CIRCLE 21 ON READER SERVICE CARD

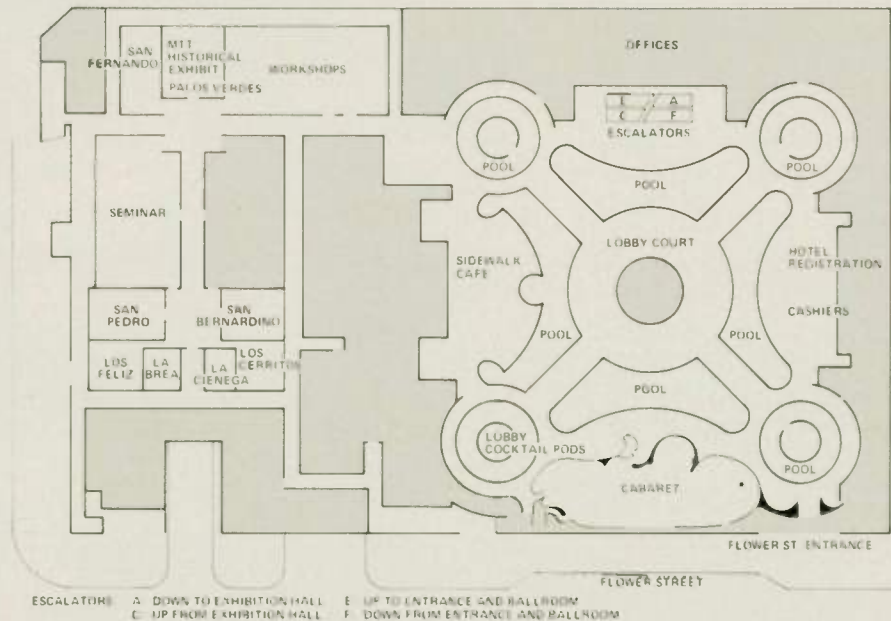
June 15-17, 1981
Bonaventure Hotel
Los Angeles, CA

**1981 INTERNATIONAL MICROWAVE
SYMPOSIUM / EXHIBITION**

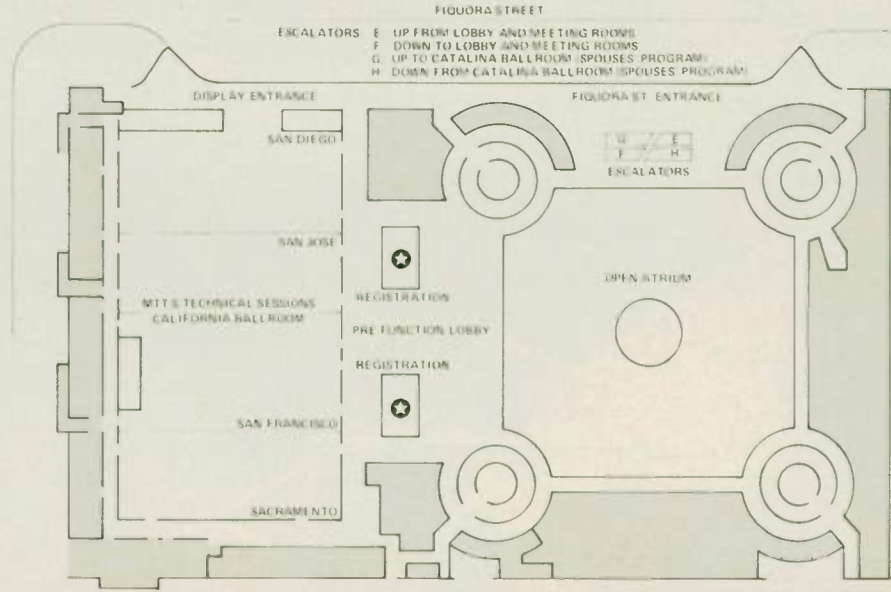
Exposition Hall (Level 2)



Lobby (Level 3)

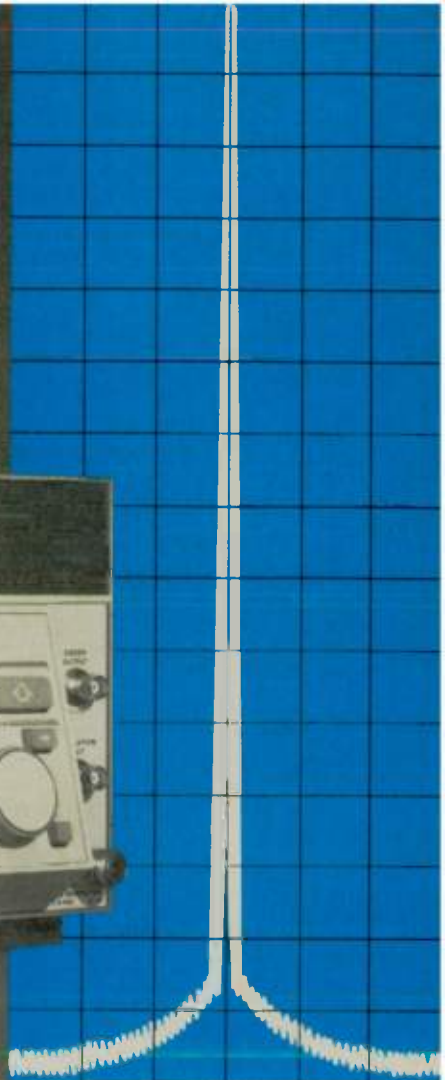


California Ballroom (Level 4)



(continued on page 32)

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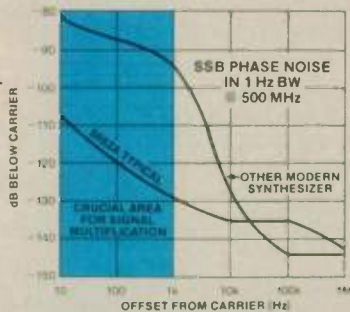
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CIRCLE 22 ON READER SERVICE CARD

at 12 GHz and up to 35% efficiency at K_u-band (paper B-1). Another paper describes 0.5 W at 21 GHz with a novel transition between microstrip and waveguide (B-2). A broadband high power FET amplifier operable from 2-18 GHz uses 600 micron gate width for 0.3 W output; a 1350 micron geometry extends this to 0.5 W in the 7-16 GHz range (B-3).

A silicon on sapphire (SOS) transceiver module design includes a power amplifier and phase shifter using MES-FET and IGFET active devices along with thin film lumped passive elements (B-5).

Session G, relating more to FET applications, contains a paper on a 4-8 GHz, 1 W, GaAs FET amplifier in a ceramic package measuring 6 x 16 x 3.5 millimeters and offering cascaded operation (G-1). This paper is followed by a 7-12 GHz, 1 W, 4 stage GaAs FET (G-2). At 13.3 GHz, a 4.5 W, 26 dB gain, 15% efficient amplifier is described having 1 dB bandwidth of 1.7 GHz (G-3). A power combiner FET amplifier produces 9 dB gain, 1.5 W, in a 6-12 GHz MIC format (G-5).

In a third session (N) GaAs Monolithic circuits are highlighted. Papers include the description of a 0.1 to 10 GHz, 4 stage low noise amplifier with active input and output matching on a 2.5 millimeter square chip (N-2). A counter operable to 1.6 GHz using a dual gate GaAs MESFET logic is described in a coplanar circuit (N-5).

Millimeter waves are highlighted in four separate sessions (C, H, N, and R). A 94 GHz, 3 dB coupler in open dielectric waveguide is described with 25 dB directivity (C-4). A W-band receiver (about 90 GHz) with 10 GHz bandwidth and 5.5 dB noise figure using 3 milliwatts LO power with zero-biased diodes is described (H-5). A 60 GHz single slot dipole radiator on etched dielectric using a built in beam lead switching diode has 10 dB of enhanced directivity (H-6). A 94 GHz subharmonic GaAs beam lead mixer has 8.5 dB double sideband noise figure and uses a 47 GHz Gunn LO and 5-500 megahertz IF bandwidth.

A 100 milliwatt, 30 GHz FET with 5 dB gain is presented in paper (M-1). An FET oscillator operable to 69 GHz has as much as 2.5 milliwatts at 57 GHz (M-2). Dual finger beam lead Schottky diodes have 4 dB noise figure at 36 GHz and usability at 95 and 140 GHz (M-3). A peak output power, of 63 W at 92 GHz using 8 IMPATT diodes in a hybrid coupled power combiner is described (M-6). A bridged double quartz standoff package is used with pulsed silicon IMPATT sources

which (used with combiners) produce 9.2 W peak at 140 GHz and 1 W peak at 217 GHz (M-7). A 9-channel down converter covering 26-60 GHz using millimeter wave IC's is described in paper (R-6). Results from testing of a 90 GHz coherent chirp radar using SAW filters and an injection locked IMPATT transmitter diode are described in paper (R-7).

Biological and medical application using microwaves are covered in papers which include a 27 megahertz waveguide application for localized hypothermia (heating) of cancerous tissue treatment (T-1). Microwave thawing of frozen packed red blood cells is described in paper (T-7).

THE AP-S & URSI TECHNICAL PROGRAM

The AP-S & URSI Symposium has been arranged to overlap (Tuesday through Friday) the MTT Conference this year at the Bonaventure. Special attention has been given to scheduling subjects of interest common to MTT Symposium participants so that they could attend (if so registered) sessions from both technical programs. No special registration is required, however, for delegates to any of the Conferences to attend the Plenary Sessions held in the San Francisco Room, Wednesday morning, 8:30 to 12 noon. The Moderator for this session, Dr. Seymour Cohn, is well known in both the microwave theory and techniques practice as well as antenna design. Four invited lectures given in this session should prove to be of keen interest to all of the technical meeting delegates. The first by T. Napier from the National Radio Astronomy Observatory, Socorro, New Mexico, is entitled "The Very Large Array Radio Telescope." The second invited lecture by S. Okwit of LNR Communications, Farmingdale, New York, is entitled "Noise and Analog and Digital Microwave Communications." This is followed by O. P. Gandhi's (from the University of Utah) "Biological Effects and Medical Applications of Electro-Magnetic Fields." The final paper is by E. C. Stone of the JPL and Physics Department, California Institute of Technology and entitled "The Voyager Encounter with Saturn." These talks can be expected to be given to a packed auditorium.

Several of the sessions specific to the URSI and AP-S programs that might be of special interest to microwave theory and techniques delegates are the sessions on Waveguides (B-2 and B-4), Microstrip (AP-S1 and AP-S12), the Array Techniques (with their implications for components) Sessions AP-S7, and AP-S16, and Dielectric Waveguides (AP-S18). ☛

ware matters is very low. The number of advanced degree candidates is in even worse shape. Where will we find the people to man the exciting programs now in hand, and those yet to come?

The roots of the problem go back a decade and more. Not only were young people turned off by technology in that era, but interest in the support of research and development by the US government and even by many companies diminished. The drying up of monies to support device and hardware related studies occurred at the time of explosive growth of the computer field, principally in the area of software. Students and faculty were quite naturally attracted to these areas.

The decay of the analog/device/hardware activities in US universities has now proceeded to the point where there is little ability and not much incentive to expand programs to supply badly needed manpower. Teaching and research lab equipment is either nonexistent or badly out of date. And student interest has eroded as well. The B.S. to M.S. to Ph.D. pipeline has been breached. There is not much incentive for a B.S. student to hold himself from the job market for four or five years. Industries have not made the salary differentials between degree levels attractive enough for a student to go to graduate school. So the seed corn gets eaten.

The perception that universities provide trained manpower as a common good has brought us to the point where companies view new engineers as a resource . . . so many needed to keep tables of organization at acceptable levels. There is little thought wasted on the production of this resource. Occasionally a company will clean out a warehouse of 10-year equipment and grant it to schools, then grumble at the quality of the product trained on 10-year-old gear.

The risks to the industry of this kind of thinking are severe. There are alternate technologies capable of displacing markets, and there is competition from around the world. This place for us, as practitioners of the microwave art, to "do the far, far better thing" and examine what we and our institutions are responsible for and how we can move to improve the level and quantity of training of those to whom leadership of our industry will pass.

On behalf of the Administrative Committee of MTT-S I'd like to congratulate Al Clavin, Don Parker and the many dedicated volunteer workers who created this year's Symposium for the benefit of us all. ☛

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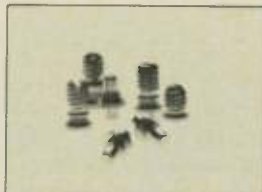


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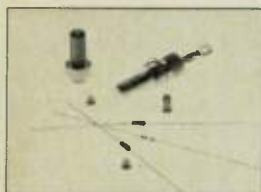
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1981 IEEE MTT-S TECHNICAL PROGRAM

MONDAY MORNING, JUNE 15, 1981

A. MILLIMETER-WAVE INTEGRATED CIRCUITS—DIELECTRIC AND IMAGE GUIDE

Chairman: H. J. Kuno

- A1 **Open Guided Wave Structures for Millimeter-Wave Circuits**
T. Itoh (Invited)
- A2 **Directive Planar Excitation of an Image Guide**
Y. Shih, J. Rivera, T. Itoh
- A3 **Slots as New Circuit Elements in Dielectric Image Line**
K. Solbach and I. Wolff
- A4 **Coupler Design in Open Dielectric Waveguide with Web Registration**
G. M. Lindgren
- A5 **Mode Conversion Effects in Bragg Reflection from Periodic Grooves in Rectangular Dielectric Image Guide**
M. J. Shiao, H. Shigesawa, S. T. Peng, A. A. Oliner
- A6 **Field Profile in a Single Mode Curved Dielectric Waveguide**
Prof. R. Mittra and T. N. Trinh
- A7 **Horn Image Guide Leaky-Wave Antenna**
T. N. Trinh, R. Mittra and R. Paleta

B. MICROWAVE FET DEVICES

Chairman: R. L. Camisa

- B1 **Sub-Half-Micron GaAs FET's for Applications through K Band**
C. Huang, A. Herbig, R. Anderson
- B2 **K-Band Power GaAs FET's**
L. S. Rosenheck, D. Herstein, I. Drukier
- B3 **2-18 GHz, High-Efficiency, Medium-Power GaAs FET Amplifiers**
H. G. Tserng, S. R. Nelson and H. M. Macksey
- B4 **Plated Source Bridged (PSB) GaAs Power FET with Improved Reliability**
T. Suzuki, M. Kobiki, M. Watase, K. Segawa and M. Irie
- B5 **Silicon-on-Sapphire (SOS) Monolithic Transceiver Module Components for L and S Band**
D. Loughton, J. Sasonoff, and J. Selin
- B6 **Optical Tuning in GaAs MESFET Oscillators**
H. J. Sun, R. J. Gutmann, and J. M. Borrego
- B7 **Microwave Model for the Dual GaAs MESFET**
G. S. F. Mau

C. COMPUTER AIDED DESIGN & MEASUREMENTS

Chairman: H. Ku

- C1 **Computer-Aided Design for the 1980s (Invited Paper)**
L. Besser, C. Holmes, M. Ball, Max Medley, S. March
- C2 **Computer-Aided Design of Microstrip Couplers with Accurate Discontinuity Models**
Erik Hammerstad

C3 Microwave Filter Design in the Time Domain

E. A. Hosney and M. I. Sobhy

C4 Microwave Analysis Using Time Domain Plots Created from Frequency-Domain Reflections

H. E. Stinehelfer, Jr. and H. E. Stinehelfer, Sr.

C5 An Interactive Optimal Postproduction Tuning Technique Utilizing Simulated Sensitivities and Response Measurements

J. W. Bandler, M. R. M. Rizk and A. E. Salama

C6 A Novel Harmonic Balancing Bridge for Characterizing Microwave Modules for Phased Array Antenna Service

Donald W. Griffin

C7 A Dual Four-Port for Automatic Analysis

H. George Oltman and Herbert A. Leach

D. LATIN AMERICAN SESSION

Chairman: A. J. Giarola

D1 Finite-Difference Method for the Arbitrary Cross-Section Waveguide Using the Best-Fit Boundary Approximation

A. B. Filho

D2 A Model of the Coupling Between Posts in Waveguide Using Equivalent Transmission Lines

E. Acosta-C., H. O. Nava-J.

D3 Slotline-Microstrip Transition on ISO-Anisotropic Substrate-Broadband Design (Invited)

A. Podcameni, M. Lima Coimbra (Invited) Analysis of Single and Coupled Striplines with Anisotropic Substrates

D4 (Invited) Analysis of Single and Coupled Striplines with Anisotropic Substrates

Adairdo Gomes d'Assuncao, Attilio Jose Giarola, David Anthony Rogers

D5 (Invited) A New Method of Pulse Dispersion Analysis for Single Mode Optical Fibers

P. S. M. Pires, D. A. Rogers, E. J. Bochove and R. F. Souza

D6 A Resonator Method for Permittivity Measurements

A. O. M. Andrade, J. J. Senise; S. S. Stuchly

D7 Microwave Modelling HF Antennas Over Lossy Earth (Invited)

Benjamin Jacurd H. and Gerardo Gomez G.

D8 (Invited) The Development of Microwave Components for Earth Station Receiver

J. K. Pinto, E. Camargo, M. Luqueze, F. S. Correra, C. A. Finardi, and E. I. Ynoue

MONDAY AFTERNOON, JUNE 15, 1981

E. MILLIMETER-WAVE INTEGRATED CIRCUITS — PRINTED CIRCUITS SESSION II

Chairman: R. T. Kihm

E1 Shielded Microstrip: Transmission Media for Millimeter Integrated Circuits

Michael Dydyk

E2 Empirical Analytical Expressions for Fin-Line Design

A. K. Sharma and Wolfgang J. R. Hoefer

E3 Experimental Assessment of Bilateral Fin-line Impedance for Device Matching

H. A. Willing and B. E. Spielman

E4 Experimental Characterization of Fin-Line Discontinuities Using Resonant Techniques

Erienne Pic and W. J. R. Hoefer

E5 A Broadband, Low-Noise Receiver at W Band

C. P. Hu and A. Denning

E6 Millimeter-Wave Planar Slot Antennas with Dielectric Feeds

P. Yen, J. A. Paul and T. Itoh

E7 94 GHz Subharmonic Mixer Using Beam Lead Diodes

J. A. Paul and P. Yen

F. FET APPLICATIONS

Chairman: R. T. Kemerley

F1 4-8 GHz High Power Cascadable Packaged GaAs FET Amplifier

S. Yamamura, M. Shigaki, N. Hidaka, H. Ishikawa

F2 Broadband Lumped-Element GaAs FET Power Amplifiers

R. L. Camisa, J. B. Klatskin, A. Mikelsons

F3 A 4.5 W, 26 dB Gain FET Power Amplifier at K_u Band

V. Sokolov and R. C. Bennett

F4 Compact Multi-Stage Single-Ended Amplifiers for S-C Band Operation

K. B. Niclas

F5 A Power FET Octave Bandwidth Traveling-Wave Combiner Amplifier

D. Kaminsky, A. G. Bert, A. Dotti

F6 Wideband Cavity Tuned GaAs FET Oscillator

Robert Joly, William W. Heinz and Edward G. Cristal

F7 Decade Bandwidth FET Functions

J. Obregon, Y. Letron, R. Funck, S. Barvet

F8 Balanced Dual Gate GaAs Frequency Doublers

Roger Stancliff

GI. THE SIX-PORT AND ITS APPLICATIONS

Chairman: S. F. Adam

GI-1 Singularities in Calibration of Six-Port Network Analyzers

H. F. Ebbesen and G. F. Engen

GI-2 The Use of a Matched Symmetrical Five-Port Junction to Make Six-Port Measurements

Gordon P. Riblet and E. R. Bertil Hansson

- G1-3 A Broadband Stripline or Coaxial 'Resolver' for the Accurate Measurement of Complex Reflection Coefficients Using the Six-Port Measurement Concept**
G. P. Riblet
- G1-4 A High-Power Dual Six-Port ANA for Determining Biological Effects of Microwave Radiation**
Cletus A. Hoer

GII. DIELECTRIC RESONATORS
Chairman: Harlan Howe

- GII-1 TM_{01} Tubular and Cylindrical Dielectric Resonator Mode**
P. Guillon, J. P. Balabaud, Y. Garault

- GII-2 Coupling of Cylindrical Dielectric Resonators to Microstrip Lines**
R. Bonetti and A. Atia

- GII-3 A Dielectric Resonator Bandstop Filter**
P. Guillon, S. Mekerta, Y. Garault

L. FERRITE APPLICATIONS
Chairman: C. R. Boyd

- L1 Coupled-Mode Theory Analysis of Distributed Nonreciprocal Devices**
I. Awai and T. Itoh

- L2 Analysis of Wideband Microstrip Circulators by Point-matching Technique**
A. M. Khilla

- L3 Performance Characteristics of Magnetoplasma Based Submillimeter-Wave Non-reciprocal Devices**
S. H. Talisa and D. M. Bolle

- L4 A Ferrimagnetic Resonance Thermometer for Microwave Power Environment**
J. A. Weiss, D. A. Hawks and G. F. Dionne

- L5 A K-Band High Power Low Loss Latching Switch**
M. J. Milnar, W. S. Piotrowski and J. E. Raue

- L6 Fin-Line Ferrite Isolator for Integrated Millimeter-wave Circuits**
A. Beyer and K. Solbach

- 15 Contiguous Broadband Matching of Multiple Resonant Loads**
J. D. Rhodes and M. J. Thornton

- 16 Inhomogeneous Broadside-Coupled Striplines**
A. G. D'Assuncao, A. J. Giarola and D. A. Rogers

- 17 Rectangular Coaxial Line Split-T Power Devices**
L. H. Yorkins

- 18 New Differential Phase Shift Networks Combining All-Pass and Band-Pass Elements**
J. K. Hunton

J. SOLID STATE CIRCUITS AND DEVICES
Chairman: C. T. Rucker

- J11 Computer-Aided Design of Microwave Parametric Frequency Dividers**
A. Lipparini, E. Marazzi, Vittorio Rizzoli

- J12 Analysis of Waveguide IMPATT Oscillator Circuits**
B. D. Bates and P. J. Khan

- J13 Coaxially Coupled Ridge Waveguide Tunable Oscillator**
R. S. Robertson, R. L. Eisenhart

- J14 High Efficiency Mode in a 20 GHz MBE IMPATT Diode Observed by a Computer-Aided Characterization System**
Hiroshi Kondoh, G. Conrad Dalman and Charles A. Lee

K. HIGH POWER CIRCUITS AND SYSTEMS
Chairman: K. Tomiyasu

(continued on page 38)

TUESDAY MORNING, JUNE 16, 1981

I. PASSIVE COMPONENTS AND NETWORKS
Chairman: H. C. Bell

- 11 (Invited) Problems in Microstrip Filter Design**
R. J. Wenzel and W. G. Erlinger

- 12 Design of Filters with Ideal Amplitude and any Prescribed Phase**
S. A. Mohammed

- 13 Miniaturized Microwave Filter Construction with Dielectric-Loaded Resonator and Space Coupling**
A. Fukasawa, T. Sato and K. Hosaka

- 14 A New Diplexer Realized in Stripline**
I. Galin

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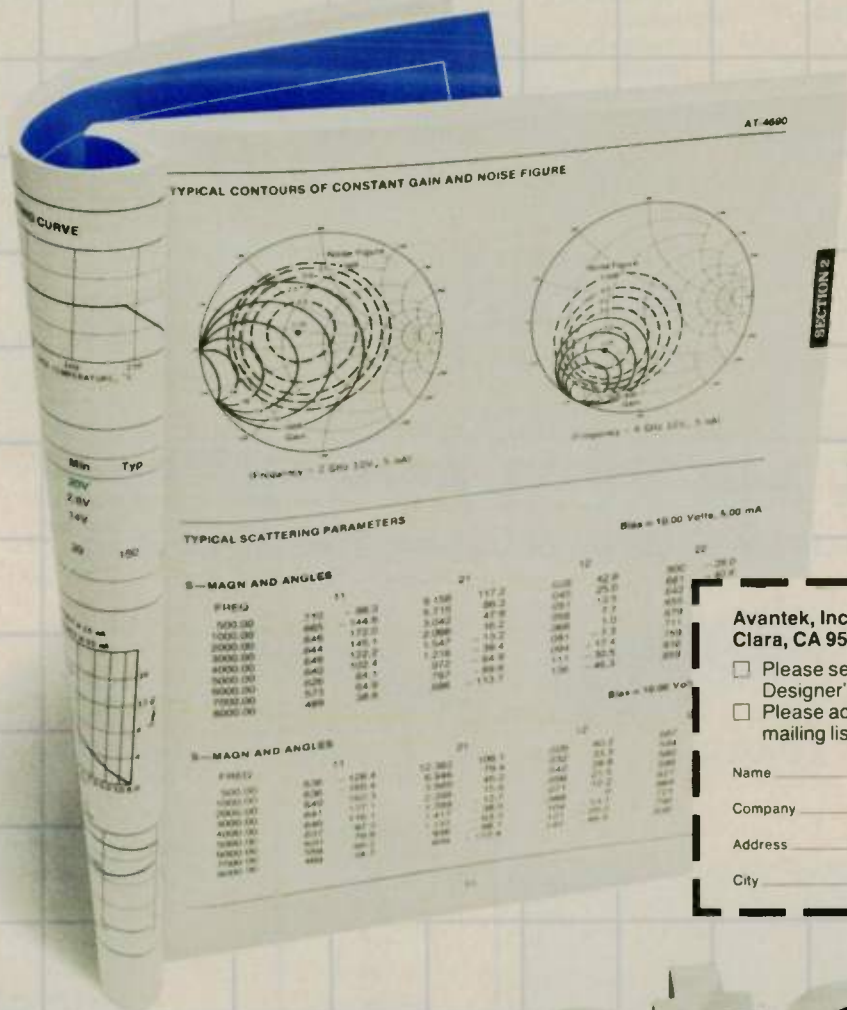
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- K1 A 100 kW Solid-State Coaxial Limiter for L Band
S. Patel and H. Goldie
 - K2 A 2 kW CW MIC 20-500 MHz SPDT Pin Diode Switch Module
R. Tenenholtz
 - K3 Development of High Power, Low Frequency Pin Diodes
M. Caulton, A. Rosen, P. Stabile and A. Gombar
 - K4 A Study of the High Power Pulsed Characteristics of Low-Noise GaAs MESFET's
D. S. James and L. Dormer
 - K5 Design of a Single-Anode, MIG-type Gyrotron Gun for a 35 GHz Gyro-TWT
J. M. Baird and A. C. Attard
 - K6 Design of a High Power Earth Station Transmitter for the Band 7.9 to 8.4 GHz
R. A. Gough
 - K7 Design and Operation of an Orotron - A Tunable Source of Coherent Millimeter-Wave Radiation
H. Dropkin, R. P. Leavitt and D. E. Wortman
 - K8 Status of the Microwave Power Transmission Components for the Solar Power Satellite
W. C. Brown
 - K9 A High Power Gyrotron Operating in the TE₀₄₁ Mode
B. Arfin and M. Read
- H. JAPANESE SESSION
Chairman: Y. Konishi
- H1 Progress of Microwave Semiconductor Devices in Japan
K. Sekido - (NFC)

- H2 (Invited)
Millimeter-wave Integrated Circuits
H. Komizo (Fujitsu)
- H3 (Invited)
Advanced RF Circuit Miniaturization for 800 MHz Land Mobile Radio Unit
S. Seki, N. Kanmuri & S. Yuki (NTT)
- H4 (Invited)
Miniaturized Diplexer for Land Mobile Communication Using Dielectric Ceramics
K. Wakino (Murata)
- H5 (Invited)
Very Low Power Gigabit Logic Circuits with Enhancement Mode GaAs MESFET
M. Omori, T. Mizutani and K. Kato
- H6 (Invited)
CH₃F Submillimeter Laser Using New Type of Resonator
M. Kawamura
- H7 (Invited)
Recent Developments of Fiber Optic Devices
K. Shirahata (Mitsubishi)
- H8 (Invited)
Optical Fiber Communication Systems in Japan
M. Yokubo and M. Kawamura (TIT)

- M1 Asymmetric Microstrip dc Blocks with Rippled Response
Darko Kajfez, S. Bokka, and C. E. Smith
 - M2 Simple Balun Coupled Mixers
Ben R. Hallford
 - M3 Computer-Aided Design of Semiconductor Mounts in Fin-Line Technology
Hadia El Hennawy and Klaus Schuenemann
 - M4 A Measurement Method for Accurate Characterization and Modeling of MESFET Chips
D. E. Peck and D. F. Peterson
 - M5 A Frequency-Stabilized MIC Oscillator Using a Newly-Developed Dielectric Resonator
Y. Komatsu, Y. Murakami, T. Yamaguchi, T. Otake, M. Hirabayashi
 - M6 A 1.75-6 GHz Miniaturized GaAs FET Amplifier Using Quasi-Lumped Element Impedance Matching Networks
S. Moghe, R. Gray, W. Tsai
 - M7 A 10.5 GHz MIC Direction Sensitive Doppler Module Using a GaAs FET and a Ag/Pd Thick Film
T. Mori, H. Sawano, K. Kusunoki, O. Ishihara
- N. MILLIMETER-WAVE SOLID-STATE DEVICES
Chairman: T. T. Fong

TUESDAY AFTERNOON, JUNE 16, 1981

M. MICROWAVE INTEGRATED CIRCUITS
Chairman: W. E. Schroeder

(continued on page 40)

RF & MICROWAVE FILTERS

LO-PASS/HI-PASS/BANDPASS/BANDSTOP/DIPLEX/SPECIALS



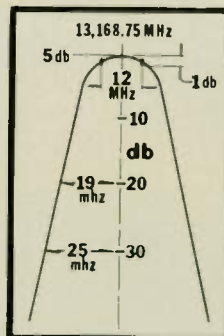
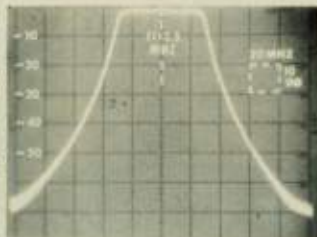
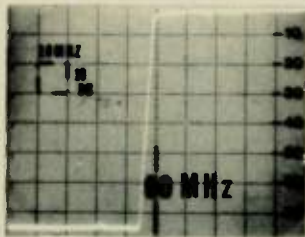
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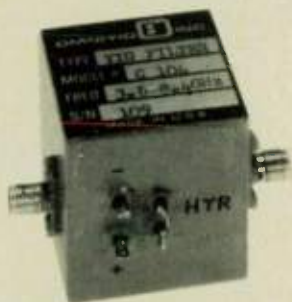
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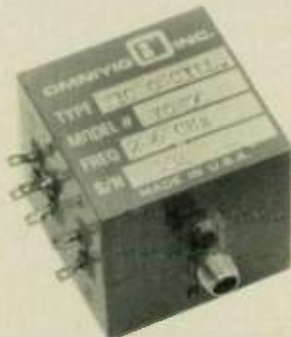
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Product Line Information Sheet



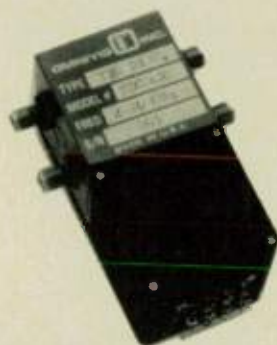
YIG FILTERS

Two stage, dual two-stage, three-stage and four-stage units. Standard octave and multi-octave designs from 0.3 to 18 GHz. Fast switching octave and multi-octave designs. Minicube sizes.



YIG OSCILLATORS

With minimum power outputs of 20, 30 and 50 mW. Standard octave bands. Fast switching. Second harmonic down greater than 20 dB. Minicube sizes. Multi-octave units have 2nd harmonic down greater than 60 dB.



YIG-TUNED DISCRIMINATORS

These units consist of a power divider, offset two-stage YIG filter and two output detectors. YIG discriminator designs are available in octave and multi-octave bands from 500 MHz to 18 GHz.

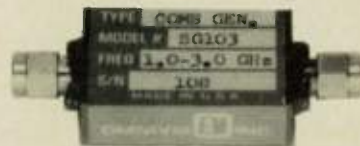


YIG-TUNED ECM FRONT ENDS

YIG-tuned integrated oscillator with pre-postselector YIG filters available in various combinations covering 300 MHz to 18 GHz in octave bands. Factory adjusted to provide user with optimum tracking, maximum RF output, lowest harmonic and spurious responses.

GUNN-EFFECT YIG OSCILLATORS

Three basic units available covering 5.9 to 6.5 GHz, 8.0 to 12.0 GHz and 12.0 to 18.0 GHz. Other bands on special order. Second harmonic down greater than 30 dB. Minimum power outputs of 40, 15 and 10 mW.



COMB GENERATORS

Standard matched comb generator designs that are self biased and having integrated elements to produce harmonics up to 18 GHz. Input is matched to 50 ohms with an integrated wide-transformer circuit which can accept a tunable input frequency. Low harmonic conversion loss and high spectrum purity.

YIG DRIVERS

Very compact and available as integral component with any YIG device. Power requirements are ± 15 volts and 0 to 10 volts for tuning the YIG device over its full frequency range. Other input voltages can be accommodated on special order.



YIG-TUNED HARMONIC GENERATORS

Step recovery diode integrated with two and three-stage YIG filter. Very compact package with high efficiency. 100 MHz input and 2.0 - 18.0 GHz output harmonics. 1.0 to 2.0 GHz continuous input and 2.0 to 18.0 GHz continuous output (minimum output at 12.4 GHz is 4 mW). 500 MHz input and 1.0 to 18.0 GHz output (1 mW at 12 GHz).



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A complete line of filters in the frequency range of 25 MHz to 18 GHz is offered. Emphasis is placed on miniaturization (such as OMNIYIG's MICRO-PAK™) while maintaining and exceeding electrical specifications. OMNIYIG also offers a complete line of equalizers to complement its line of filters.



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Narrow band, octave and multi-octave detectors 0.1 to 18 GHz designs in Tunnel, Schottky or Zero Bias Schottky diode types. Pin and Schottky Limiters for octave and multi-octave bands 0.1 to 8 GHz ranges. Many package styles available. Special requirements welcome.

- N1 **A 30 GHz 100 mW GaAs FET**
Yoshinobu Kadowaki, Osamu Ishihara,
Masao Nakatani
- N2 **A 69 GHz FET Oscillator**
J. M. Schellenberg, H. Yamasaki,
D. W. Maki
- N3 **Beam-Lead Schottky-Barrier Planar
Mixer Diodes for Millimeter-Wave
Applications**
S. Jamison, A. Contolatis, P. Bauhahn,
M. Helix and J. Abrokwah
- N4 **A Wideband, Backshort Tunable
Second Harmonic W-Band Gunn
Oscillator**
H. Barth
- N5 **Metal-Barrier-Metal Junctions for
Room Temperature Millimeter-Wave
Mixing and Detection**
C. W. Slayman and T. K. Gustafson
- N6 **A 63 W W-Band Injection-Locked
Pulsed Solid State Transmitter**
H. C. Yen and K. Chang
- N7 **Millimeter-Wave Solid State Devices**
Kai Chang, F. Thrower and
G. M. Hayashibara
- N8 **Chip Level IMPATT Combining at
40 GHz**
C. T. Rucker, J. W. Amoss and
G. N. Hill
- O. **GaAs MONOLITHIC CIRCUITS**
Chairman: G. Chao

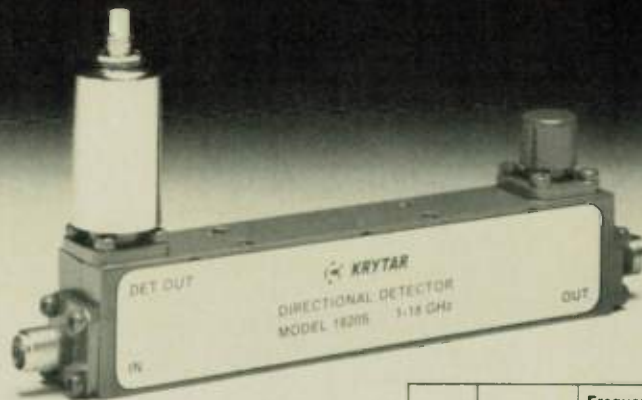
- O1 **A Study of Optional Matching Circuit
Topologies for Broadband Monolithic
Power Amplifiers**
J. E. Degenford, D. C. Borre,
R. G. Freitag and M. Cohn
- O2 **A Monolithic GaAs 0.1 to 10 GHz
Amplifier**
W. C. Peterson, D. R. Decker,
A. K. Gupta, J. Dully, D. R. Chen
- O3 **A 2 12 GHz Feedback Amplifier or
GaAs**
K. B. Niclas and W. T. Wilson
- O4 **A Planar Type Low-Noise GaAs
Monolithic Microwave Amplifier**
Deng Xian can and Zhu Guo-liang
- O5 **A High-Speed Monolithic GaAs
10/11 Counter**
R. E. Lundgren, D. E. Synder
- O6 **An 8 GHz MMIC Preamplifier**
D. R. Decker, A. K. Gupta,
W. C. Peterson and D. R. Chen
- O7 **A Comparison Between Actively and
Passively Matched S-Band GaAs Mono-
lithic FET Amplifiers**
J. R. Suffolk, R. S. Pengelly, Jr.,
R. Cockrill and J. A. Turner
- O8 **Use of Switching Q in the Design of
FET Microwave Switches**
H. A. Atwater and R. W. Sudbury

- P. **MICROWAVE ACOUSTICS**
Chairman: T. Lukaszek
- P1 **SAW Based Direct Frequency
Synthesizer**
Allan Budreau
- P2 **SAW Stabilized Radiosonde**
D. J. Dodson, K. F. Lau, M. Y. Haung
and T. J. Lukaszek
- P3 **SAW Oscillators UHF Transit Satellite**
B. Y. Lao, N. J. Schneider, D. A. Rowe,
R. E. Dietterle, J. S. Schoenwald,
E. J. Staples and J. Wise
- P4 **Tunable Magnetostatic Surface Wave
Oscillator at 4 GHz**
R. L. Carter, J. M. Owens,
W. R. Brinlee, Y. W. Sam and
C. V. Smith, Jr.
- P5 **IC Compatible SAW Device on GaAs
(Invited)**
T. Grudkowski
- P6 **SAW Bandpass Filter Components for
Microwave Systems**
D. E. Allen, F. S. Hicknerell
- P7 **A SAW Interferometer Direction-
Finding and Frequency Identification
Method**
Dirk Klose, William Skudera
- P8 **Magnetostatic Wave Compressive
Receiver**
C. E. Nothnick, J. F. Billing,
M. R. Daniel, and T. D. Adams

(continued on page 42)

BROADBAND DIRECTIONAL DETECTORS

.5-18 GHz — with Unequaled Flatness

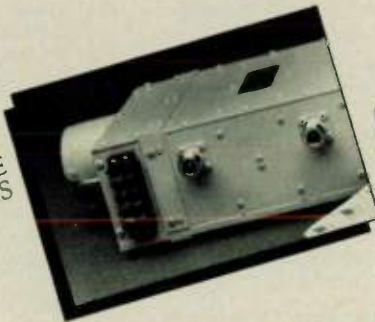


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Model	Frequency (GHz)	Frequency Sensitivity (dB) (GHz)	Directivity (dB) (GHz)	Max VSWR	Sensitivity (μV/μW)	Price
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1818S	2-18	± 5 2-12.4 ± 7 2-18	17 2-12.4 15 12.4-18	1.35	10	\$750
1820S	1-18	± 5 1-12.4 ± 7 1-18	17 1-12.4 15 12.4-18	1.35	10	\$825
1850S	5-18	± 1.2	14 5-18 12 12.4-18	1.40	10	\$925

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DIGITALLY CONTROLLED TUNABLE RF FILTERS



CONTIGUOUS BAND MULTIPLEXERS



FILTERS, FILTERS, FILTERS



Do you need to: Eliminate cross talk in installations with antennas in close proximity?; Isolate and identify signals coming in to wide-band receiver systems?; Preselect frequencies in automatic test equipment? Is a tunable filter the last remaining piece of manual equipment in your digital system?

Telonic's new digitally addressable tunable rf filters can solve these problems. They're offered in 22 standard models (3 or 5 sections) covering 11 frequency bands between 48 MHz and 4 GHz. Each capacitively loaded Chebyshev filter model tunes over a 2:1 frequency range for full octave coverage.

Integration of the filter into your digital system is made easy by a wide choice of standard interfaces, such as RS-232, TTL serial, and IEEE STD-488 (GPIB) parallel.

The combination of time-proven, rugged tunable filter technology with state-of-the-art microprocessor-based control results in a highly stable, repeatable system which can't drift, requires no daily recalibration, and has been extensively field-tested in hostile environments.

Call us with your tunable filter needs. Small quantities can be delivered in as little as one week. Send for new technical data sheet.

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The multiplexers are 0.1 dB ripple Chebyshev type with housings milled from a single aluminum block for package integrity and mechanical stability.

A typical design may have passbands of 100 to 200 MHz, 200 to 400 MHz, and 400 to 1000 MHz. The upper section can be either a bandpass or highpass filter, and the lower section can be either a bandpass or lowpass filter—to fit your exact needs.

The units have a typical insertion loss of 0.2 to 0.3 dB with a crossover insertion loss less than 5.0 dB.

CIRCLE 9

If you're looking for an RF or microwave filter in the 20 MHz to 12 GHz range, we probably have a model to match your exact application.

We offer more than 23 types of filters, including highpass and subminiature series, and the new computer-controlled tunable RF filter. We also offer tubular lowpass, tubular bandpass, cavity interdigital bandpass, tunable cavity bandpass, combline, miniature and tunable band reject filters.

Our new filter catalog (shown below) provides an easy-to-use filter selection guide, frequency and bandwidth tolerance curves, passband relationships and performance curves for each filter type. Send for your free copy.

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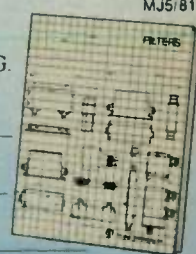
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WEDNESDAY, JUNE 17, 1981

PLENARY SESSION JOINT IEEE AP-S, MTT-S, URSI

Moderator: Dr. Seymour Cohn

1. The Very Large Array Radio Telescope
Dr. Peter Napier
2. Noise and Analog and Digital Microwave Communications
S. Okwit
3. Biological Effects and Medical Applications of Electro-magnetic Fields
O. P. Gandhi
4. Voyager Encounter with Saturn
Dr. E. C. Stone

WEDNESDAY AFTERNOON, JUNE 17, 1981

- Q. MICROWAVE AND MILLIMETER-WAVE SYSTEMS
Co-Chairmen: H. Sobol, J. B. Horton
- Q1 A Phase Alignment Network for Space Diversity Combining
G. L. Heiter, H. Miedema
- Q2 Direct Generation of MSK Modulation at Microwave Frequencies

S. Kumar, W. J. Chudobiak and J. S. Wight

- Q3 A Dielectric Resonator Filter as Low-Loss Delay Element for 14 GHz On-Board 4-phase PSK Demodulation
L. Accatino, A. Angelucci
- Q4 14-GHz Differential QPSK Demodulator for Regenerative Satellite Repeater
G. Ohm, M. Albery, and D. Rosowsky
- Q5 Channelized Receiver Covering 26 to 60 GHz with Planar Integrated Circuit Components
P. J. Meier, K. D. Breuer, L. D. Cohen, N. Worontzoff, J. Lepore and J. Gunther
- Q6 A W-Band Coherent Pulse-Compression Radar Transceiver Using Linear Frequency Modulation
C. Brenneise, M. Beebe, T. Kihm and R. D. Weglein

R. PHASED AND ACTIVE ARRAY TECHNIQUES

Chairman: R. Van Wagoner

- R1 A Scanning Switch Matrix for a Cylindrical Array
K. J. Keeping, D. S. Rogers and J. C. Sureau
- R2 Beam Steering Antenna Control Techniques
A. R. Skatvold
- R3 35 GHz Active Aperture
M. F. Durkin, R. J. Eckstein,

M. D. Mills, M. S. Stringfellow and R. A. Neidhard

- R4 The Use of IMPATTs in Active Arrays
S. Hamilton, G. Stern
- R5 Broadband Dual-Gate GaAs FET Continuously Variable Phase Shifter
M. Kumar, H.C. Huang and R. J. Menna
- R6 Active Microwave Power Combiner/Divider Using Dual Gate MESFET
J. J. Pan
- R7 A Continuously Variable K_u -Band Phase/Amplitude Control Module
Y. Gazit, H. C. Johnson

THURSDAY MORNING, JUNE 18, 1981

S. LOW NOISE TECHNIQUES

Chairman: S. Okwit

- S1 A Low Noise Solid State Amplifier for Replacement of a K_u -Band TWTA
P. H. Wolfert, J. D. Crowley, F. B. Fank
- S2 Cooled Low Noise GaAs Monolithic Mixers at 110 GHz
B. J. Clifton, G. D. Alley, R. A. Murphy, W. J. Piacentini, I. H. Mroczkowski and W. Macropoulos
- S3 A 4 GHz Low Noise GaAs FET Amp
Li Hao-Mo
- S4 Impact of Low Noise Technology on Present and Future Satcom Systems (Invited)
H. C. Okean

(continued on page 44)



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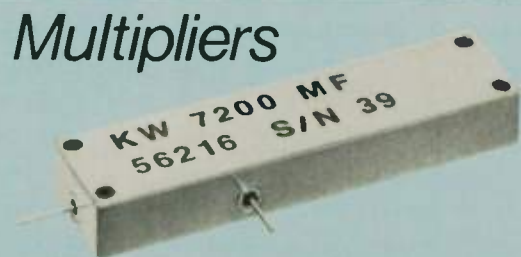
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Frequency range	5 - 18 GHz	0.5 - 12 GHz	95 - 13.3 GHz	0.1 - 18 GHz
Comb Spacing (MHz)	500	500	950	100
Freq. Stability (ppm)	30	30	30	30
Power Output (dBm)	-40 min	+6 max -22 min	-20 min	0 max -40 min
Load VSWR (max)	1.5:1	1.5:1	1.5:1	1.5:1
In-band Spurious (dBc max)	>24	>30	>40	>60
Power Supply	+15 Vdc at 200mA max	+15Vdc at 200mA max	+15Vdc at 150mA max	+15Vdc at 300mA max
Size (inches nominal)	1.9 x 1.33 x .5	1.33 x 1.33 x .5	1.5 1.9 x 1.33 x .06	1.9 x 1.33 x .06



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Freq. Range (GHz)	MODEL #		Isolation dB Min.	Insertion Loss dB Max.	VSWR Max.
	Circulators	Isolators			
1.0 - 2.0	50A1021	60A1021	16	0.6	1.35
2.0 - 4.0	50A3001	60A3001	18	0.5	1.30
2.6 - 5.2	50A3011	60A3011	18	0.5	1.30
4.0 - 8.0	50A6001	60A6001	18	0.5	1.30
5.0 - 10.0	50A6071	60A6071	18	0.5	1.30
8.0 - 12.4	10B9201	20B9201	20	0.4	1.30
8.0 - 16.0	50A2001	60A2001	17	0.5	1.35
12.0 - 18.0	10B2201	20B2201	18	0.5	1.30
8.0 - 18.0	50A2051	60A2051	16	0.6	1.10

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OSCILLATORS

DIELECTRICALLY STABILIZED OSCILLATORS (DSO)

Model Number	Frequency Range (GHz)	Output Power (mw Minimum)	Size, inches nominal
6600 - 1300	3.65 - 4.05	30	2.7 x 1.7 x 1
6600 - 1610	4.25 - 4.75	30	2.6 x 1.5 x .9
6600 - 1611	4.75 - 5.25	30	2.5 x 1.5 x .9
6600 - 1612	5.10 - 5.60	30	2.5 x 1.5 x .9
6600 - 1613	5.40 - 5.90	30	2.5 x 1.5 x .9
6600 - 1614	5.60 - 6.10	30	2.5 x 1.5 x .9
6600 - 1615	5.70 - 6.20	30	2.5 x 1.5 x .9
6600 - 1616	5.90 - 6.40	30	2.5 x 1.5 x .9
6600 - 1617	6.40 - 6.90	30	2.5 x 1.5 x .9
6600 - 1618	7.00 - 7.45	15	2.2 x 1.3 x .9
6600 - 1910	8.00 - 8.40	15	2.2 x 1.3 x .9
6600 - 1911	8.60 - 9.00	15	2.2 x 1.3 x .9
6600 - 1912	9.00 - 9.50	15	2.2 x 1.3 x .9



FREQUENCY STABILITY: 10 ppm/°C max
5 ppm/°C, typical

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- S5 Highly Reliable Low Noise MM-Wave Mixers with Whisker-Contacted Honeycomb Diodes
Joerg Schroth
- S6 A Subharmonically Pumped Fin-Line Mixer for Satellite TV Receiver Applications
Gunther Begemann
- T. BIOLOGICAL EFFECTS AND MEDICAL APPLICATION
Chairman: W. R. Adey
- T1 27 MHz Waveguide Applicators for Localized Hyperthermia Treatment of Cancer
R. W. Paglione, F. Sterzer, J. Mendecki, E. Friedenthal and C. Botstein
- T2 Microstrip Loop Radiators for Local Hyperthermia
I. J. Bahl, S. S. Stuchly, J. W. Lagendijk and M. A. Stuchly
- T3 A New Optical Technique for the Measurement of Temperature in RF and Microwave Fields
K. A. Wickerhshem and R. V. Alves
- T4 Non-Perturbing Temperature Probe and Thermography Measurements in Microwave Diathermy
G. Kantor and C. Hochuli
- T5 Clinical RF Hyperthermia (Invited)
F. Krishan Storm
- T6 The Effects of High Power Microwave Pulses on Red Blood Cells and the Relation to Transmembrane Thermal Gradients
Susan L. Gartner, A. W. Friend, K. R. Foster and H. Howe, Jr.
- T7 Microwave Thawing of Frozen Packed Red Blood Cells
Nancy L. Campbell; J. Drewe
- T8 Application of Moment-Methods to Electromagnetic Biological Imaging
M. J. Hagmann, O. P. Gandhi, D. K. Ghodgaonkar
- T9 Microwave Imaging: Numerical Simulation and Results
M. F. Ishkander, R. Maini, C. H. Durney
- U. MICROWAVE FIELD AND NETWORK THEORY
Chairman: S. March
- U1 Propagation Parameters of Coupled Microstrip-like Transmission
S. K. Koul, B. Bhat
- U2 Hybrid Mode Analysis of Microstrip Lines on Anisotropic Substrates
A. M. El-Sherbiny
- U3 Analysis of Microstrip Line on Semiconductor Substrate
M. Anbourg, J. P. Villotte, F. Godon and Y. Garault
- U4 Wave Propagation in Inhomogeneous Anisotropic Rectangular Waveguides by the Effective Index Method
M. N. Armenise, M. De Sario
- U5 Improved Technique for Evaluation of Slot Discontinuities in Rectangular Waveguide
P. K. Park, G. J. Stern
- U6 Desegmentation of Two-Dimensional Microwave Circuits and its Application to Stripline Power Dividers
K. C. Gupta, P. C. Sharma and R. Chadha
- V. GUIDED WAVE OPTICS AND SIGNAL PROCESSING
Chairman: J. K. Butler
- V1 The Integrated Optical Spectrum Analyzer — A First Demonstration
T. R. Raganath, J. Lee and T. Josephf
- V2 Guided Wave Optical RF Spectrum Analyzer (Invited)
D. Mergerian, E. C. Malarky, R. P. Pantemis, J. C. Bardley and A. L. Kellner
- V3 A Generalized Two-Dimensional Coupled-Mode Analysis of Curved and Chirped Periodic Structures in Open Dielectric Waveguides
Zong-Q. Lin, Shu-Tong Zhou, William S. C. Chang, S. Forouhar and J. Delavaux
- V4 Nonreciprocal Propagation Characteristics of YIG Thin Film
Tetsuya Mizumoto, Yoshiyuki Naito
- V5 Experiment on Light Intensity Modulation Based on Guided-to Radiation Mode Coupling in Hetero-Structure Thin Film Waveguide
H. Onodera, I. Awai, M. Nakajima and J. Ikenoue
- V6 Polarization-Rotated Radiations Conversion in Electrooptic Waveguides
S. Yamamoto

(continued on page 46)

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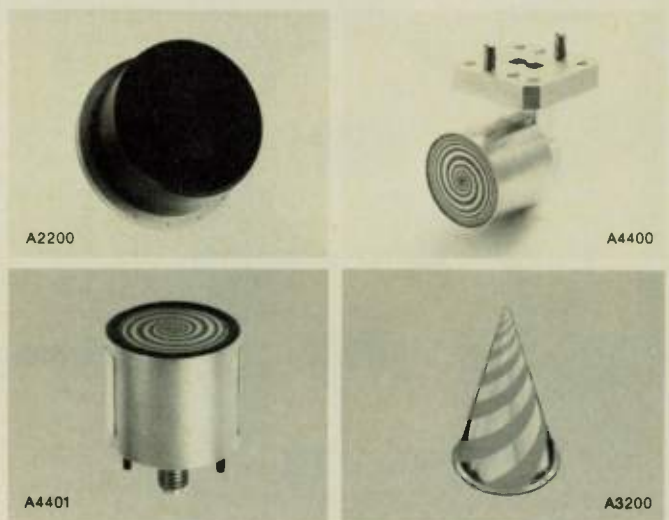
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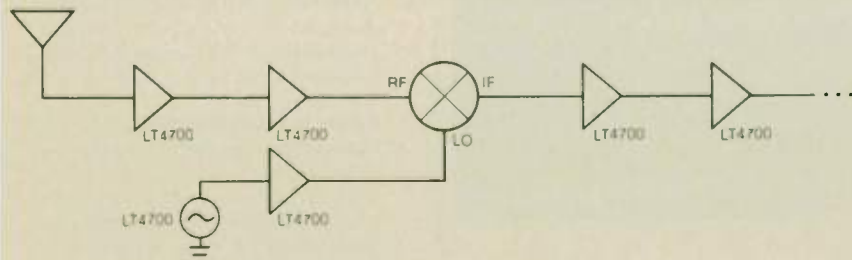
Company

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State Zip



*Noise measure M_{min} is the system noise figure of an infinite cascaded chain of identical amplifier stages

$$M_{min} = 10 \log_{10} \left(1 + \frac{NF_{min} - 1}{1 - 1/G_{ANF}} \right) \text{ where } NF_{min}, \text{ minimum noise figure, and } G_{ANF}, \text{ gain at associated noise figure, are specified as power ratios.}$$

FOR TRW's LT4700, $M_{min} = 1.22 \text{ dB}$ at 0.5 GHz and 1.64 dB at 1.0 GHz

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CIRCLE 33 ON READER SERVICE CARD

1981 IEEE/AP-S & URSI TECHNICAL PROGRAM

TUESDAY MORNING, JUNE 16, 1981

AP-S SESSION 1

MICROSTRIP ANTENNAS (II)

Chairman: Prof. S. Long

1. Resonant Frequency of Rectangular Microstrip Antennas, M. C. Bailey and M. D. Deshpande
2. A Wide Band, Multiport Theory for Thin Microstrip Antennas, W. F. Richards, and Y. T. Lo
3. Cylindrical Rectangular Microstrip Antenna Radiation Efficiency Based on Cavity Q Factor, C. M. Krowne

4. Comparison of Radiation Characteristics of Microstrip Patch Antennas, I. J. Bahl and P. Bhartia
5. Segmentation and Desegmentation Techniques for Analysis of Planar Microstrip Antennas, K. C. Gupta and P. C. Sharma
6. A Simple Accurate Formula for the Radiation Conductance of a Rectangular Microstrip Antenna, H. Pues and A. Van de Capelle
7. Characteristic of Microstrip Ring

8. Cross-Polarized Radiation from Metal Patch Type Microstrip Antenna Elements, Donald W. Griffin

URSI SESSION B-1 SCATTERING I

Chairman: Dr. G. A. Thiele
Commission B

1. Electromagnetic Scattering by Open Circular Waveguides, T. W. Johnson, and D. L. Moffatt
2. Scattering from Open Cylinders, L. N. Medgyesi-Mitschang, and C. Eftimiu
3. Diffraction by a Perfectly-Conducting Plane Angular Sector: Improvements on the Eigenfunction Solution and Some Numerical Results, J. N. Sahalos, and G. A. Thiele
4. Plane Wave Diffraction by a Wedge - A Spectral Domain Approach, Clarkowski and R. Mittra
5. New Asymptotic Wedge Diffraction Coefficients, A. Mohsen
6. Interaction of a Plane Electromagnetic Wave with a Dielectric Body: A New Hybrid Iterative Solution, R. Bansal, T. T. Wu and R. W. P. King
7. Boundary Integral Equations for Scattering from Complex Dielectric Structures, R. E. Kleinman, N. J. Damaskos and J. R. Jameson
8. Reflection from Gyroelectromagnetic Layered Media, N. J. Damaskos, A. L. Maffett and P. L. E. Uslenghi
9. High Frequency Diffraction from a Strip in a Flat Plane with Arbitrary Impedance Boundary Conditions, R. Tiberio and G. Pelosi

AP-S SESSION 2

ARRAY PATTERN SYNTHESIS

Chairman: Dr. W. A. Imbriale

1. A New Approach to the SNR Optimization of Lossy Receiving Arrays, G. Y. Delisle, M. G. Pelletier, J. A. Cummins and J. Munier
2. Theory of Null Synthesis of Planar Arrays, Sharad R. Laxpati and J. Paul Shelton
3. Applications of Null Synthesis to Hexagonal Arrays, J. Paul Shelton and Sharad R. Laxpati
4. Performance of Thinned Antenna Arrays, D. S. Hicks and D. C. Patel
5. Elimination of the Dummy Elements in Thinned Phased Arrays, F. Beltran, and F. King
6. Approximate Methods of Solution for Element Patterns in Thinned Arrays, F. Beltran
7. A General Theory of Weighted Phased-Array Antenna and the Analysis of Phase Weighting, Fang Neng hang

URSI SESSION B-2 WAVEGUIDES I

Chairman: Prof. S. T. Peng
Commission B

1. Axisymmetric TM Modes in Open Cylindrical Waveguides with Central

(continued on page 48)

FREQUENCY (MHZ)	INPUT RETURN LOSS (S11) DB	RETURN ANG	GAIN FORWARD (S21) DB	ANG (S21) ANG	ISOLATION REVERSE (S12) DB	ANG (S12) ANG	OUTPUT RETURN LOSS (S22) DB	RETURN ANG (S22) ANG
10.000	-27.1	79.2	12.09	-176.3	-41.00	14.8	-21.4	124.6
25.000	-27.7	7.7	12.21	176.2	-40.68	16.1	-26.8	118.2
50.000	-28.0	-40.7	12.25	167.3	-39.59	23.9	-30.7	108.8
75.000	-28.2	-66.1	12.28	159.2	-38.20	29.5	-33.7	92.6
100.000	-28.1	-83.4	12.31	151.2	-36.79	32.1	-36.3	68.8
125.000	-27.8	-97.6	12.32	143.3	-35.48	31.8	-36.7	-32.1
150.000	-27.0	-110.1	12.34	135.5	-34.26	29.5	-34.7	-56.6
175.000	-25.8	-121.9	12.35	127.5	-33.19	26.5	-32.9	-75.4
200.000	-24.9	-133.5	12.36	119.5	-32.22	23.1	-31.3	-90.7
225.000	-24.0	-145.4	12.37	111.6	-31.36	19.2	-29.8	-106.3
250.000	-23.1	-156.9	12.37	103.4	-30.58	15.2	-28.2	-123.3
275.000	-22.5	-167.8	12.37	95.5	-29.89	10.6	-26.9	-137.4
300.000	-21.7	-178.5	12.37	87.4	-29.26	5.9	-25.5	-148.4
325.000	-21.0	-170.6	12.37	79.2	-28.71	1.3	-24.6	-159.4
350.000	-20.4	160.7	12.37	71.3	-28.20	-3.5	-23.9	-172.9
375.000	-19.8	152.2	12.30	63.3	-27.76	-8.7	-23.8	173.3
400.000	-19.3	143.2	12.30	55.2	-27.05	-13.6	-24.1	161.9
425.000	-18.8	133.6	12.28	47.3	-26.76	-18.7	-24.1	150.3
450.000	-18.5	125.1	12.28	39.3	-26.50	-23.8	-23.9	133.1
475.000	-18.1	117.3	12.27	31.3	-26.30	-29.0		
500.000	-17.3	108.4	12.30	23.1				

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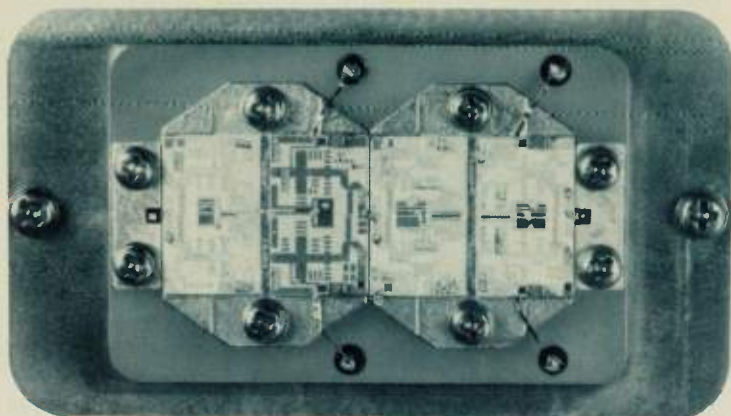
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CGB596403-08	35	0	236x85x40	15V/3.3A
CGB596405-09	37	0	318x85x40	15V/4.8A
CGB596410	40	0	—	15V/10.2A

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FUJITSU MICROELECTRONICS

(from page 46) TECHNICAL PROGRAM

- Conductor, G. C. Sherman and C. Hennessy
2. Analysis of a Tunnel Waveguide Junction, D. A. Hill
 3. Mode-Conversion Effects in Radiation from Periodically-Corrugated Dielectric Structures, A. A. Oliner and S. T. Peng
 4. Quasi-Optics of the Coupling of Guided Modes in a Corrugated Dielectric Waveguide, S. R. Seshadri
 5. Mode Filters for Integrated Optics: Application of Anisotropic Dielectrics, B. B. Chaudhuri and D. K. Paul
 6. Spectral Domain Analysis of Discontinuity Microstrip Structures, A. K. Sharma, and B. Bhat
 7. An Experimental Study of Nonuniform Surface Reactances Synthesized Using a Grounded Pin-Bed Structure, K. S. Park, D. V. Thiel, R. J. King and W. S. Park
 8. Some Applications of Non-Uniform Waveguides, N. Berkane and J. Ch. Bolomey
 9. Modes of an Array of Dielectric Waveguides, Leonard Eyges

AP-S SESSION 3

HORN FEEDS

Chairman: Dr. S. J. Hamada

1. Corrugated Waveguide Monopulse Feed, P. J. B. Clarricoats, and R. Elliot
2. Cross-Polar Behavior of Wide-Angle Corrugated Horns, P. J. B. Clarricoats, S. Mahmoud and A. D. Oliver
3. High Performance Dual-Band Corrugated Feed for TVBS Antenna for the Italian Region, S. Ghosh, N. A. Adatia, B. K. Watson
4. Control of Modal Amplitudes in Overmoded Waveguides, N. Sridhar, T. K. Sen and G. P. Srivastava
5. A Circular Horn with a Dielectric Coated Flange as a Feed for Prime Focus Paraboloidal Reflectors, A. Ittipiboon, L. Shafiq, E. Bridges and F. Rahman
6. Radiation from Coaxial Cavity Feed with Aperture Reactance (IRIS) Loading, Yeong-ming Hwang, Levent Ersoy, and Lawrence F. Brokish
7. An X-Band Single Horn Autotracking Antenna Feed System, T. Yodokawa and S. J. Hamada

URSI SESSION F-1

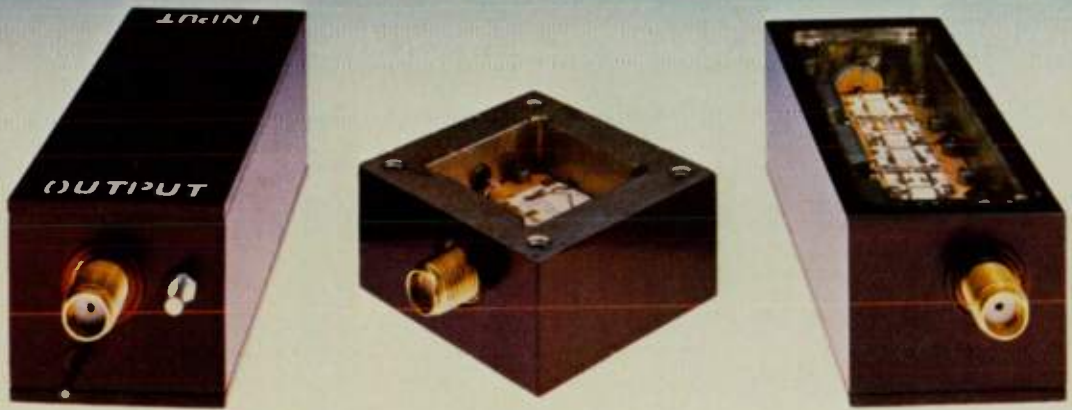
REFRACTION EFFECTS

Chairman: Dr. C. Crain

Commission F

1. A Simplified Model for Interpreting the Doppler Spectrum of Forward Scatter Radar Signals, R. E. Post and H. M. Ibrahim
2. A Study of Troposcatter Signal Behaviour, N. C. Mathur, P. R. Pande, R. K. Bassi and V. K. Girdhar
3. Improving Knowledge of Multipath Fading, L. F. Mojoli
4. A Technique for Obtaining Tropospheric Refractive Range Error Corrections from Angle Error Measurements, E. E. Altshuler and K. Mano
5. On the Maxwell-Equation Basis for Radiative Transport Theory, R. L. Fante

(continued on page 50)



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	SDA 80180-05	8 - 18	24	7.5	+10	+20
Medium power	SDA 2040-13	2 - 4	38	6.5	+21	+30
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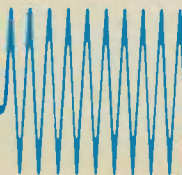
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(from page 48) TECHNICAL PROGRAM

- Analytic Solution for the Two-Frequency Mutual Coherence Function for Spherical Wave Propagation, D. L. Knepp
- On Field Computation in Radio Frequency Interferometry, A. Mohsen
- Convergence Characteristics of Geophysical Tomography Using Iterative Ray Tracing, R. D. Radcliff, C. A. Balanis, H. W. Hill, Jr. and G. E. Trapp, Jr.

AP-S SESSION 4

SCATTERING AND MEASUREMENTS

Chairman: Dr. Cleyon O. Yowell

- Microwave Backscatter Measurements on Transmission Lines, D. L. Odom and N. C. Currie
- Millimeter-Wave Backscatter of Transmission Cables, H. H. Al-Khatib and J. Fielding
- RCS Measurements of Cavities, R. B. Dybdal and H. E. King
- Characterization of the Scattering Properties of an Antenna by Rapid Swept-Frequency Measurements, Johnson J. H. Wang
- Rapid Broadband Swept-Frequency Automated Scattering Measurement, C. D. Papanicolopoulos, R. L. Moore, and J. J. H. Wang
- Radar Antenna/Mast Near-Field Interactions, Werner Koppl
- Field Analysis of a Class of RF Intrusion Sensor Systems, Harold R. Raemer
- High Sensitivity Millimeter-Wave Instrumentation, R. B. Dybdal, T. T. Mori and H. E. King

AP-S SESSION 5

APERTURE AND LENS ANTENNAS

Chairman: Dr. W. C. Wong

- High Efficiency V-Band Fan Beam Antenna, C. C. Chen
- A New Method for Waveguide Lens Analysis, L. Grun and S. W. Lee
- Broadband Astigmatic Compensation, T. S. Chu
- Vectorial and Scalar Approaches for Determination of Interbeam Isolation of Multiple Beam Antennas — A Comparative Study, Y. Rahmat-Samii and A. B. Salmassi
- A New Approach to Random Aperture Errors, Victor K. Tripp
- Ray Structure Method for Coupling Coefficient Analysis of the Two Dimensional Rotman Lens, Michael J. Maybell
- Aperture Field Nodes Generated by a Butler Matrix for the Design of Very Low Side Lobe Electronic Scanning Reflector Antennas, Shii-Shyong Wang and Hang-Kam Izaguirre
- The Solution to the Problem of the Circular Aperture Distribution with One Parameter in Antenna Design, Yu Kuan Mao and Kaizhou Chen

TUESDAY AFTERNOON, JUNE 16, 1981

URSI SESSION B-3

SCATTERING II

Chairman: Prof. R. E. Kleinman
Commission B

(continued on page 53)



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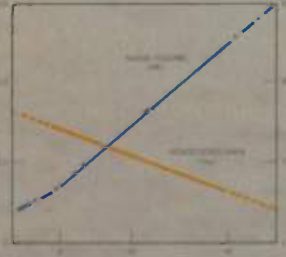
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1. A Simple Technique for Solving E Field and H-Field Integral Equations for Conducting Bodies at Internal Resources, T. K. Sarkar
2. On Improving the Stability of the Electric Field Integral Equation at Low Frequencies, D. R. Wilton and A. W. Glisson
3. A Hybrid Technique Which Combines the Moment Method with an Asymptotic Current, T. J. Kim, and G. A. Thiele
4. An Improved Mode-Matching Method for the Problem of Scattering by Conducting Cylinders with Arbitrary Shape, M. Nishimura, S. Takamatsu and H. Shigesawa
5. On the Scattering from an L-Shaped Wire, C. D. Taylor
6. Electromagnetic Diffraction by a Narrow Slit in an Impedance Sheet — E-Polarization, D. F. Hanson
7. Comparisons Between Measurements and Computations of HF Radar Backscatter from Canonical Ship-Like Structures, E. K. Walton and E. H. Newman
8. Scattering Area of a Large Wind Turbine Blade, D. L. Sengupta and J. E. Ferris
9. Experimental Determination of Electromagnetic Energy Absorption on Complex Shapes; A Progress Report, R. M. Sega, V. M. Martin, and R. W. Burton

**IRSI SESSION B-4
WAVEGUIDES II; OPTICAL AND
MILLIMETER WAVEGUIDES**

Chairman: Dr. E. F. Kuester

Commission B

1. Propagation Constants for Step-Index Optical Fibers of Arbitrary Cross-Section at High Frequency, E. F. Kuester and R. Ebrahimian
2. On the Analysis of Inhomogeneities in Optical Waveguides, T. C. Kaladhar Rao
3. Optical Communications with Elliptical Fibers, R. Y. Wong and S. R. Rengarajan
4. Investigating the Properties of the Radially Inhomogeneous Optical Fibers Using the Predictor-Corrector Method, H. H. Yao and G. L. Yip
5. A Numerical Solution for Inhomogeneous Dielectric Waveguides, Ch. Pichot
6. Scattering of Surface-Waves by a Dielectric Step Discontinuity, H. Shigesawa and M. Tsuji
7. Suspended and Inverted Microstrips for Millimeter Wave Applications, J. Rivera and T. Itoh
8. New Polarization Preserving Single-Mode Fibers — The Layered Fibers, C. Yeh, and A. R. Johnston

AP-S SESSION 6

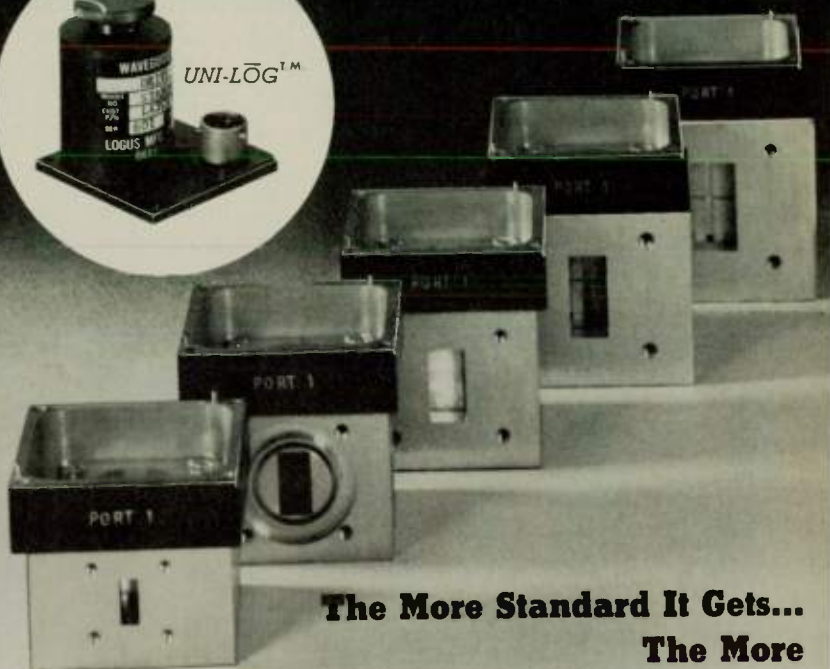
SIGNAL PROCESSING ANTENNAS

Chairman: Dr. J. B. Damonte

1. Radar Waveform Synthesis for Target Discrimination, Kun Mu Chen, Dennis P. Nyquist, Doug Westmoreland, Che-I Chuang and Byron Drachman
2. Correlation of Masquerading Multimode, Keith A. Struckman
3. Effective Pattern and Anti-Clutter Performance of Frequency Scanning Radar Antennas, Ji Sheng Zhong

(continued on page 54)

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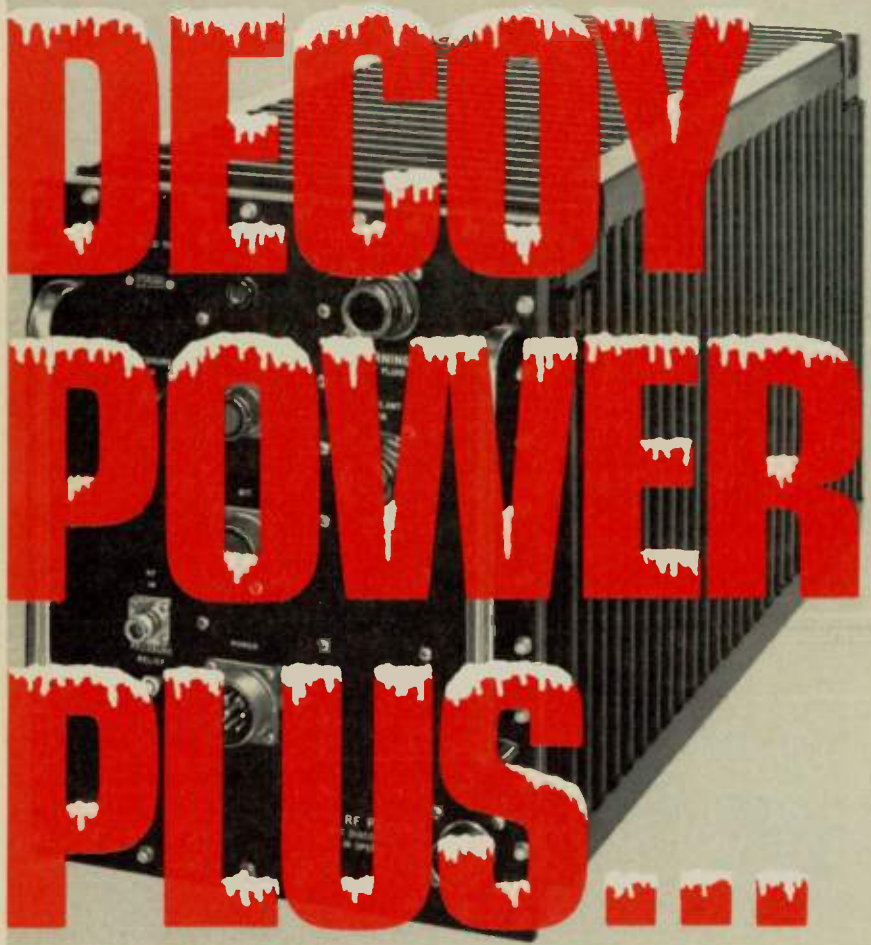
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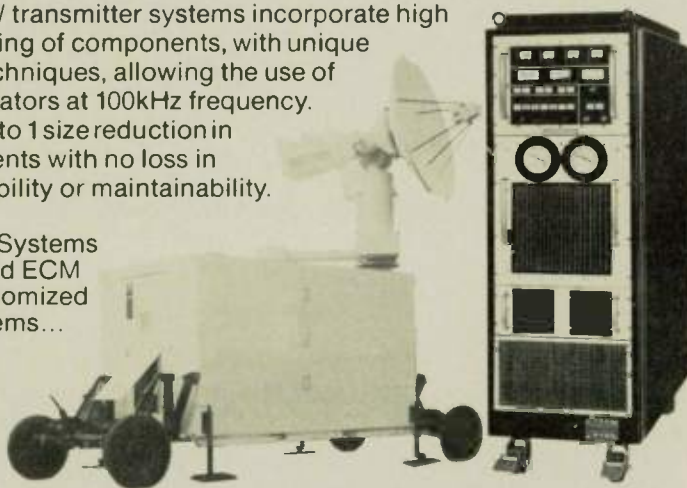


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(from page 53) TECHNICAL PROGRAM

4. Direction-Finding Using a Discrete Space-Matched Filter, J. Munier, G. Y. Delisle, J. A. Cummins, and B. Faure
5. Comparison of Signal-to-Noise Ratio in the UHF Band for Antennas with Linear and Circular Polarization, Victor Lander
6. Electric Field Induced Rate Effects in Pharmacokinetic Systems, Y. J. Seto and S. T. Hsieh
7. Analysis of the Time Domain Aperture Antenna with Theory of Linear System, Shi-ming Lin
8. Time Domain Analysis of the Transient Behavior of an Aperture Antenna, Shi-ming Lin

AP-S SESSION 7 ARRAY TECHNIQUES

Chairman: Dr. Giorgio Borgiotti

1. The SHF Hybrid Scan Array, A. J. Kelly, E. Pinck, J. F. Pedersen and W. W. Bogin
2. A Wideband Shaped-Beam Low-Sidelobe Conformal Array Step-Scannable via a Modified Blass Network, P. J. McVeigh and R. M. Rudish
3. Near Field Transmission Lines Feed Linearly Polarized Array Antenna for Built-In Test, Shii-Shyong Wang and Ruey-Shi Chu
4. Radiation Performance of a Large Slotted Waveguide Array for Satellite Borne Radiometric Applications, Mostafa Afifi Ray Meier, and Louis Dod
5. Sidelobe Requirements for the Microwave Landing System, Alfred R. Lopez
6. A Simple Coaxial FED Planar Microstrip Slot Array, F. Rahman, L. Shafar, E. Bridges, and A. Ittipiboon
7. Kevlar Armored Phased-Array, S. Y. Peng, R. S. Chu, N. S. Wong and R. Tang
8. Large X-Band Electronically Scanned Array, D. E. Bostrom, J. A. Castaneda, R. E. Hodges and P. Fujii

AP-S SESSION 8 HORN ANTENNAS

Chairman: Dr. P. G. Ingerson

1. Characteristics of a Square Coaxial Waveguide Feed for a Reflector Antenna, Sharad R. Laxpati, C. Q. Lee, and Craig M. Skarpiak
2. A GTD Parametric Study of Pyramidal Horns for Reflector Applications, J. Huang, Y. Rahmat-Samii, and K. Woo
3. A Wideband Low-Sidelobe Disc-O-Cone Antenna, J. L. Wong and H. E. King
4. Surface Wave Launched and Scattered on the Discontinued Impedance Surface and the Radiation Theory of the Surface Wave Antenna, Ri-rong Zhang and Xiao-Jie Gao
5. Mode Coupling Coefficients in Conical Horn Junction, V. Daniele, M. Orefice and R. Zich
6. An Octave Bandwidth Meanderline Polarizer Consisting of Five Identical Sheets, Derek A. McNamara

IPS SESSION 9**ANTENNA MEASUREMENTS (II)**

Chairman: Prof. Edmond S. Gillespie

- 1. Verification Testing of a Spherical Near-Field Algorithm and Comparison to Compact Range Measurements, Doren W. Hess and Joseph J. Tavormina
- 2. Spherical Scanning Data Processing; An Algorithm for Halving the Data Processing Effort When the Radiation into the Back Hemisphere is Negligible, Richard Lewis
- 3. A Second-Generation Near-Field Range, C. P. Burns, V. V. Jory, D. F. Tsao, R. B. Cotton, E. N. Bone and A. N. Morrison
- 4. A Highly Accurate Planar Near-Field Scanner, F. E. Shannon and J. L. Luzwick
- 5. Near-Field Measurements of Broad Beam Antennas, T. B. Wells
- 6. Spherical-Wave Source-Scattering-Matrix Analysis of the Mutual Coupling Between Two Antennas, Richard Lewis
- 7. An Improved Design of the Swept-Frequency Automatic-Z-Measuring Schemes Using Multiple Probes, Chia-Lun J. Hu
- 8. The Gain Measurement of Large Antennas with Cosmic Radio Sources at Frequencies above 10 GHz, Noriyuki Kawaguichi and Noakazu Hamamoto

IPRSI SESSION F-2**ATMOSPHERIC ATTENUATION, SCATTERING, AND ABSORPTION**

Chairman: Dr. R. T. Woo

Commission F

- 1. Attenuation and Depolarization Over an Earth-Space Path at 12 GHz: Measurement Results for Path Elevation Angles of 27 and 9 Degrees Using the CTS Satellite, A. J. Rustako, Jr.
- 2. Measured Bounds on Rain-Scatter Coupling Between Space-Earth Radio Paths, D. C. Cox, H. W. Arnold and H. H. Hoffman
- 3. Fade Duration and Interfade Interval Statistics Measured on a 19 GHz Earth-Space Path, H. W. Arnold, D. C. Cox and H. H. Hoffman
- 4. A Site Diversity Experiment Using the Sirio Spacecraft, C. W. Bostian, R. E. Marshall, W. L. Stutzman and W. P. Overstreet
- 5. Transmission and Backscattering of Optical Waves through Fog, R. L-T Cheung and A. Ishimaru
- 6. Application of Satellite Microwave Radiometry in Meteorology, N. C. Grody
- 7. Some Results of HF Radio Noise Measurements in Nigeria Using Directional Antenna, C. O. G. Obah
- 8. Crosspolarization of Ice Particles at 30 GHz, C. Yeh, R. Woo and J. W. Armstrong, and A. Ishimaru
- 9. Microwave Depolarization and Rain-drop Canting Angles, R. S. Butler
- 10. Earth-Space Rain Attenuation Measurements at 13 GHz for Several Sites in Canada, R. V. Webber, J. I. Strickland and J. J. Schlesak

(continued on page 56)

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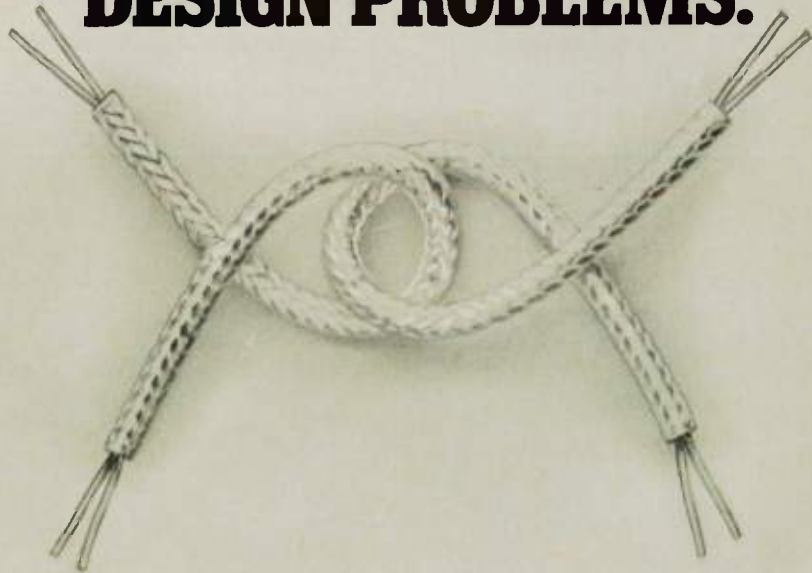
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WEDNESDAY MORNING, JUNE 17, 1981

PLENARY SESSION
JOINT IEEE AP-S, MTT-S, URSI

Moderator: Dr. Seymour Cohn, Consultant

1. The Very Large Array Radio Telescope, Dr. P. Napier
2. Noise and Analog and Digital Microwave Communications, Dr. S. Okwit
3. Biological Effects and Medical Applications of Electromagnetic Fields, Dr. O. P. Gandhi
4. The Voyager Encounter with Saturn, Dr. E. C. Stone

WEDNESDAY AFTERNOON,
JUNE 17, 1981

SESSION B-5
GOUBAU MEMORIAL SESSION

Commission B

Organizer: David C. Chang

Session Chairman: James W. Mink

Eulogy: Felix Schwering

Leo Felsen, Chairman International

Commission B on Fields and Waves

1. Theory of Beam Waveguides: A Review, F. Schwering
2. Some Aspects of Georg Goubau's Experimental Investigations at Fort Monmouth, J. R. Christian and J. W. Mink
3. Mathematical Model of Multi-Element Monopole Antennas, G. Goubau (deceased), N. N. Puri and F. Schwering
4. Guiding Mechanisms on Open, Planar Structures, D. C. Chang, and T. Itoh
5. Radiation Characteristics of Leaky Dielectric Waveguides, R. Mittra
6. Bending Losses of Open Waveguides, E. F. Kuester
7. New Physical Effects Due to Mode Coupling in Various Dielectric Structures, A. A. Oliner and S. T. Peng

AP-S SESSION 10
ADAPTIVE ANTENNAS

Chairman: Prof. J. G. Fikioris

1. Adaptive Antenna/Receiver-Processor System, G. G. Chadwick, J. G. Charitat, W. Gee, C. C. Hung and J. L. McFarland
2. A UHF Communication Antenna Employing Open Loop Control Null Steering, D. E. N. Davies and J. R. F. Guy
3. Maximally Flat Second Order Feedback Loops for Fast Adaptive Antenna Arrays, I. A. Mandour, S. E. El-Khany and M. A. Abouldahb
4. Design Limitations on Adaptive Array Control Loop Nulling Time, Lawrence R. Burgess, and Dr. Raymond S. Berkowitz
5. Weight Dynamic Range in Power Suppression Adaptive Arrays, Heng Cheng Lin

6. Cross Coupled Boot Strapped Interference Canceler, Yeheskel Bar-Ness, and J. Rokach
7. Angle of Arrival Performance Bounds in an Interfering Environment, E. and A. Mendelovicz and E. T. Oestreich
8. Angular Tracking Results for a Highly-Quantized Adaptive Array, Lloyd J. Griffiths and Charles W. Jim

AP-S SESSION 11
NUMERICAL METHODS AND FORMULATIONS

Chairman: Prof. K. Mei

1. Scattering and Absorption of Lossy Dielectric Objects Irradiated by the Near Fields of Aperture Source, A. Lakhtakia, M. F. Iskander, C. H. Durney and H. Massoudi
2. Cutoff in Fin Lines Evaluated with the Spectral Domain Technique and with the Finite Element Method, A. K. Sharma, G. I. Costache and W. J. R. Hoefler
3. A Spectral-Iteration Technique for Analyzing Scattering from Arbitrary Bodies, Raphael Kastner and Raj Mittra
4. A Generalized Numerical Expression for the Three-Dimensional Laplacian Operator, E. E. Okon
5. Effect of the MFIE Resonance on the Approximate Solution of the Reaction Matching Integral Equation Applied to a Rotationally Symmetric Antenna, A. Hizal, K. Aydin and F. Canatan
6. Application of the Boundary Element Method to Electromagnetic-Scattering Problems, M. H. Lean and A. Westlar
7. Electromagnetic Waves on Cylindrical Structures Calculated by the Method of Moments and by the Point-Matching Technique, C. Hafner and R. Ballisti
8. A Delta-Distribution Derivation of the Electric Field in the Source Region, Arthur D. Yaghjian

AP-S SESSION 12
MICROSTRIP ANTENNAS (III)

Chairman: Prof. T. Itoh

1. A Perturbation Approach to the Design of Circularly Polarized Microstrip Antennas, Y. T. Lo, and W. F. Richards
2. Microstrip Backfire Antenna, C. M. Kalor, D. Hatfield and P. Simon
3. Temperature Compensation of Microstrip Antennas, M. A. Weiss
4. Input Impedance Smith Chart Curves for H-Plane Mutual Coupling Between Two Rectangular Microstrip Antennas, C. W. Krowne, and A. R. Sinduris
5. The Impedance of an Elliptical Printed-Circuit Antenna, Stuart A. Long and Mark W. McAllister
6. Printed Circuit Antenna for Wide Bandwidth Requirements, Kenneth D. Arkind and Richard L. Powers
7. Broadbanding of Microstrip Antennas by Orthogonal Polarizations, Sazanami Yano, and Akira Ishimaru
8. Expanding the Frequency Bandwidth of a Microstrip Antenna, Yasuo Suzuki, Noriaki Miyano and Taneaki Chiba



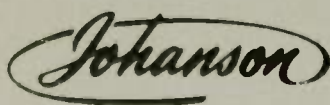
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**URSI SESSION H-1
LABORATORY, SPACE AND FUSION
PLASMAS**

Organizer and Session Chairman: Prof. N. C. Luhmann, Jr.

Commission H
INVITED PAPERS

1. **Magnetic Field Line Reconnection Experiments**, R. L. Stenzel and W. Geikelman
2. **Plasma Waves in Saturn's Magnetosphere**, F. Scharf
3. **Recent Theoretical and Experimental Progress in Heating in the Vicinity of Cyclotron Harmonics of Fusion Plasmas**, J. Scharer
4. **Lower Hybrid Heating**, T. Yamamoto
5. **Particle Beam Experiments in Space**, D. Papadopoulos
6. **Ion Bernstein Waves: Plasma Heating, Ion Temperature Diagnostics and Current Drive Feasibility**, M. Ono

**URSI SESSION B-6
RADAR WAVEFORMS AND
IDENTIFICATION**

Chairman: Prof. A. A. Ksienski

Commission B

1. **Radar Spectroscopy: T-Matrix Approach**, P. J. Moser, and H. Uberall
2. **R-Matrix Theory of Radar Scattering**, P. J. Moser; J. D. Murphy, A. Nagl and E. Uberall
3. **Synthesis of Radar Signal for Exciting a Single-Mode Backscatter from a Sphere**, K-M Chen and D. Westmoreland
4. **An Approach to Radar Target Identification**, S. W. Lee, R. Mittra and M. Hurst
5. **Modelling of Nonlinear Scatterers with Nonlinear Radar Cross Sections**, J. Y. Hong, Y. C. Kim and E. J. Powers
6. **Hypothesis Testing Applied to Radio Direction Finding**, S. Rashba and I. Kohlberg
7. **An Automated Microwave Measurement Facility for Three-Dimensional Tomographic Imaging by Wavelength Diversity**, N. H. Farhat and C. W. Werner

**URSI SESSION B-7
INVERSE SCATTERING**

Chairman: Prof. P. L. E. Uslenghi
Commission B

1. **An Inverse Scattering Method for One-Dimensional Problems in Absorbing Media**, A. K. Jordan
2. **Direct and Inverse Scattering of Combined Fields**, P. L. E. Uslenghi
3. **An Algorithm for Profile Inversion**, C. Q. Lee
4. **On Some Aspects of Microwave Imaging**, J. Ch. Bolomey, A. Izadneghadar and M. Solaimani
5. **Characterization of a Cylindrical Shell. A Geometrical Optics Approach to an Inverse Scattering Problem**, B. Duchene, and W. Tabbara
6. **Inverse Scattering - Inverse Source Theory**, N. N. Bojarski

7. **An Exact, Closed-Form Solution for the Source Term in the Inverse Source/Inverse Scattering Problem, and an Analytic Solution for the Scattering Potential**, W. R. Stone
8. **Unique, Well-Posed, and "Real World" Considerations for an Analytic, Closed-Form Solution to the Inverse Scattering Problem**, W. R. Stone
9. **The Application of the Exact, Closed-Form Solution to the Inverse Scattering Problem to the Remote Probing of Inhomogeneous Media**, W. R. Stone

**AP-S SESSION 13
PROPAGATION**

Chairman: Prof. Akira Ishimari

1. **Fading Statistics on a 23 km Link at 9.6 and 28.6 GHz**, R. H. Ott
2. **Antenna Beamwidth Independence of Measured Rain Attenuation on a 28 GHz Earth-Space Path**, H. W. Arnold, D. C. Cox and H. H. Hoffman
3. **Transequatorial Propagation on VHF and Low UHF Bands During the Peak of Solar Cycle 21 for the Euro-African Sector**, C. Fimerelis and N. K. Uzunoglu
4. **Scattering of a Plane Wave from Cylindrical Tunnels Inside a Lossy Medium**, N. K. Uzunoglu and J. D. Kanellopoulos
5. **Scattering of Electromagnetic Plane Wave from a Slightly Random Surface**, H. Ogura, J. Nakayama and M. Sakata
6. **An Experimental Study of VLF Waveguide Propagation**, Fu Yaozong

**AP-S SESSION 14-A
TRANSIENTS**

Chairman: Dr. Jeremy Landt

1. **Wideband Radar Target Response Modeling**, H. Mieras and C. L. Bennett
2. **Transient and Broadband Analysis of Aircraft Fixed-Wire Antennas**, Lennart Marin
3. **A Preprocessing Technique for Identification of Scatterer Resonances from Measured Transient Response**, J. R. Auton and M. L. Van Blaricum

**AP-S SESSION 14-B
SEM**

Chairman: Dr. Carl Baum

1. **Source-Synthesis Considerations in SEM Equivalent Circuits for Energy-Collecting Structures**, Krzysztof A. Michalski and L. Wilson Pearson
2. **Distortion Analysis of Non-Linearly Loaded Antennas**, C. Naldi and R. Zich
3. **A Modified Time-Domain Inverse Scattering Scheme in Presence of Random Noise and System Distortions**, Sujeet K. Chaudhuri and Paul A. Lenz
4. **Some Physical Constraints on the Use of Nonsinusoidal Electromagnetic Waves Radiation**, Kaihsienk Jen

**AP-S SESSION 15
SLOTTED ANTENNAS**

Chairman: Mr. D. K. Waingo

1. **Attenuation in Slot Antennas Due to Waveguide Wall Thickness**, R. W. Lyon and A. J. Sangster
2. **Design of Transverse Slot Arrays Fed by Boxed Stripline**, P. K. Park and R. S. Elliott

(continued on page 58)

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3. An Evanescent-Mode Waveguide Frequency-Scanning Feed, R. I. Wolfson, C. F. Cho and G. A. Hockham
4. Millimeter-Wave W-Band Slotted Waveguide Antennas, F. G. Farrar
5. Effect of Baffles on the Mutual Admittance Between Slot Antennas, N. G. Alexopoulos, P. J. Contoyannis and G. Franceschetti
6. Near Field Investigation of Coupling Between Two Horns, Letritia V. Muresan and T. J. F. Pavlese
7. Fields at the Aperture of Pyramidal Horns, S. I. Ghobrial

AP-S SESSION 16-A
PHASED ARRAY ANALYSIS
 Chairman: Dr. N. Amitay

1. Radiation from Infinite Arrays of Arbitrarily Shaped Thin Wire Antennas, H. K. Schuman, D. R. Pflug and L. D. Thompson
2. Performance of a Two-Dimensional Microstrip "Patch" Element in a Phased Array, C. C. Liu, J. Shmoys and A. Hessel
3. Transform Feed for Low Sidelobe Space-Fed Lens Phased Array Antenna, Jerome D. Hanfling and Bradley R. Herrick
4. Narrow Multibeam Satellite Ground Station Antenna Employing a Linear Array with a Geosynchronous Arc Coverage of 60°, N. Amitay and M. J. Gans

AP-S SESSION 16-B
CONFORMAL ARRAYS
 Chairman: Dr. Y. Rahmat-Samii

1. A Conical Array of Dielectric Filled Edge Slot Antennas, A. R. Sindoris and S. T. Hayes
2. Mutual Coupling in Rotationally Symmetric Arrays of Circular Apertures on Conducting Spherical Surfaces, P. K. Bondyopadhyay and A. Hessel
3. Conical Phased Array Antenna Investigations, A. T. Villeneuve and W. H. Kummer
4. A Low-Cost Lightweight C-Band Cylindrical Array, K. J. Keeping, D. S. Rogers and J-C. Sureau

AP-S SESSION 17
REFLECTOR ANTENNAS - I
 Chairman: Dr. C. C. Chen

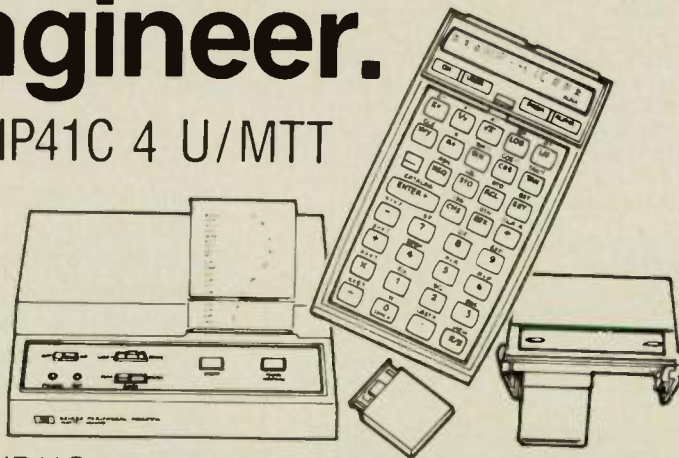
1. The Use of GTD in the Design and Analysis of Low Sidelobe Reflector Antennas, Yeogming Hwang and C. C. Han
2. Dual Polarized Elliptical Beam Antenna for Satellite Application, N. Adatia, B. K. Watson and S. Ghosh
3. Radiation Pattern Prediction for Arbitrary Reflector Antennas, M. C. Bailey, and Steven M. Moskowitz
4. Sidelobe Level Reduction by Improvement of Strut Shape, Naota Matsunaka; Shinichi Betsudan, and Takashi Katagi
5. A Compact and Low Sidelobe Offset Antenna for 4/5/6 GHz Band Terrestrial Communication Links, Nobuo Nakajima and Yoshitaro Shimanuki
6. Reduction of the Strut Radiation of Reflector Antennas, Herbert Thielen
7. A Rigorous Approach to the Theory of Paraboloidal Antenna, Lang Jen

URSI SESSION F-3
SENSING SURFACES
 Chairman: Dr. E. K. Smith
 Commission F

1. Surface-Current Solution for Rough-Surface Scattering, J. C. Leader
2. Remote Sensing of Ocean Surface Current and Current Shear by Microwave Radar, D. L. Schuler
3. Polarization Effects in Scattering from Breaking Waves, L. B. Wetzel
4. Emission from an Irregular Layer Embedded with MIE Scatterers, A. K. Fung, K. K. Lee and H. J. Eom
5. Radiative Transfer Theory for Microwave Remote Sensing of an Inhomogeneous Medium with Rough Surfaces, R. T. Shin, S. L. Chuang and J. A. Kong
6. Random Medium Model for Microwave Remote Sensing of Earth Terrain, J. A. Kong, R. T. Shin, F. J. Vallese, S. L. Chuang and J. K. Lee
7. Ocean Surface Area Coverage by Satellite-Mounted Radars, R. E. McIntosh and R. S. Raghavan
8. Another Look at Ocean Surface Current Measurements from Satellites, C. T. Swift and R. E. McIntosh

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**JRSI SESSION B-8
TRANSIENTS AND SEM**

Chairman: L. Bennett

Commission B

1. Volterra Series Solution of Maxwell Equations in Nonlinear Media, G. Franceschetti and I. Pinto
2. The Natural Frequencies of the Axisymmetric TE Modes of a Hollow Conducting Sphere with a Circular Aperture, K. F. Casey
3. Pole Extraction in the Frequency Domain, J. M. Pond and T. B. A. Senior
4. Physical Basis of the Singularity Expansion Method in Terms of Creeping Waves (Theory and Experiment), H. Uberall, A. Nagl and J. V. Subrahmanyam; B. Z. Hollmann; G. C. Gaunard, and J. D. Murphy
5. Practical Solutions to Transient Electromagnetic Probing of Stratified Lossy Media, J. Ch. Bolomey, D. Lesselier and G. Perronnet
6. Identification of Scatterer Resonances by Reversing the Measured Unforced Response in Time, M. L. Van Blaricum and J. R. Auton
7. Data Acquisition and Processing in Transient Scattering Measurements, M. A. Morgan and C. W. Hammond
8. Transient Measurements of Linear Antenna Arrays in Free Space and Over Perfectly Conducting Ground, K. D. Reich, K. J. Langenberg, P. Fellingner and K. Mayer

**JRSI SESSION B-9
NUMERICAL AND ANALYTIC
TECHNIQUES**

Chairman: Dr. D. R. Wilton

Commission B

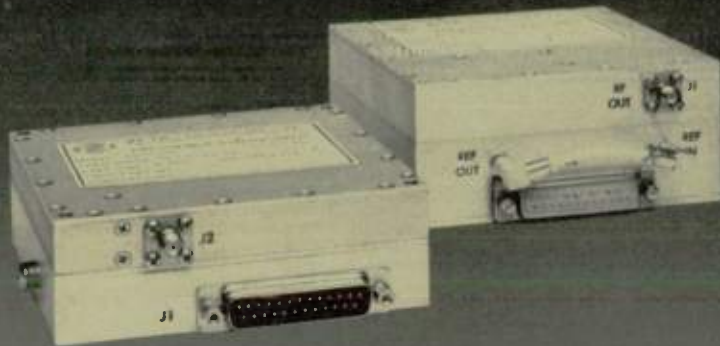
1. Optical Wavelength Calculation of Scattering Cross Sections with the Numerical Electromagnetic Code (NEC), B. A. Blevins
2. Quadrifilar Helix Antenna Design for Satellite Communications Using the Numerical Electromagnetic Code, D. E. Fessenden
3. A New Local File Manipulation Technique for Solving the Scattering of Large Objects, C. D. Papanicolaopolos and J. J. H. Wang
4. Two Hybrid Techniques for the Treatment of Wire Antennas Near an Elliptical Cylinder, K. Siakavara and H. Papadimitrakaki-Chichlia; J. N. Sahalos, and G. A. Thiele
5. A Hybrid FD-TD Approach to Electromagnetic Wave Backscattering, A. Taflov and K. Umashankar
6. FFT Utilization in Moment Method Solutions Using Entire Domain Bases, M. Teward and A. Q. Martin
7. A Moment Solution of the Diffusion Equation Using Measured Boundary Conditions, N. J. Damaskos
8. The Best Least Squares Rational Approximation of Transfer Functions by Solution of a Linear Uncoupled Eigenvalue Problem, T. K. Sarkar

1. Quasi-Optics of Magnetic Waves Guided by a Planar Film, S. R. Seshadri
2. Cut Off Wavenumbers of the Symmetrical Modes for an Eccentric Cladded, Three Layer, Dielectric Waveguide, N. Metrou, J. D. Kanellopoulos, J. A. Roumeliotis and J. G. Fikioris
3. On Plane Wave Propagation in Composite Dielectrics, Benjamin Rulf and Herbert Feldman
4. Analysis of Propagation Characteristics of Light Waves in a Periodically Corrugated Dielectric Film by Perturbation Method, N. S. Chang and Y. Matsuo

1. Theory of Dielectric Grating Antennas of Finite Width, S. T. Peng and A. A. Oliner, and F. Schwing
2. A Modified Normal Mode Helix with High Impedance and Bandwidth, Mario Orefice
3. The Circular Loop Antenna Near a Material Interface, L. N. An and G. S. Smith
4. Improved 243 MHz Homing Antenna System for Use on Helicopters, F. Klinker

(continued on page 60)

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**AP SESSION 19-A
RADIATION AND E/M RESONANCE**

Chairman: Dr. Ken Casey

1. **Group and Representation Theory of Finite Groups and Its Application to Field Computational Problems with Symmetrical Boundaries**, C. Hafner and P. Leuchtman
2. **Characteristics of Double Aperture Cavity Resonances**, C. H. Liang and D. K. Cheng
3. **Inductive Shielding of Single- and Multi-Layered Enclosures - A Review**, K. S. H. Lee
4. **Radiation Fields of Metallic Prolate Spheroid Excited by Any Circumferential Slot**, Lang Jen and Chuan-Shui Hu

**AP-S SESSION 19-B
ANTENNA MEASUREMENTS (II)**

Chairman: Mr. R. I. Wolfson

1. **Proposed Method to Eliminate Errors in Antenna Pattern Measurements Due to Reflections**, Jose Perini
2. **Range Distance Requirements for Measuring Low and Ultralow Sidelobe Antenna Patterns**, P. S. Hacker and H. E. Schrank
3. **Measurement of Low Cross-Polarized Patterns**, Donald G. Bodnar
4. **Synthesis and Evaluation of Spherical Quasi Plane Wave Regions for Antenna Pattern Measurements**, Jorgen Hald

**AP-S SESSION 20
LOW FREQUENCY ANTENNAS**

Chairman: Prof. C. A. Balanis

1. **A Frequency Tracking, Tuned, Receiving Monopole**, Gerome Reeve and Arthur Wannwright
2. **A Compact Wideband HF Transmitting Antenna**, Richard K. Royce
3. **Broadband Trapped Multiple-Wire Antennas**, W. A. Edson
4. **Analysis of a V-Loop Antenna**, Naomi Amano, and Risaburo Sato
5. **Decoupling of a Monopole Antenna with a Radial-Transmission-Line Choke**, T. G. Dalby
6. **Monopole Antenna Patterns on Finite Size Composite Ground Planes**, Constantine A. Balanis, and Dennis DeCarlo

**URSI SESSION B-10
ANTENNAS I: REFLECTORS AND RADOMES**

Chairman: Dr. Y. Ranmat-Samii

Commission B

1. **Some Limitations in the Broadband Operation of Segmented Reflectors**, J. M. Tranquilla, P. J. Sutherland and R. H. Joyce
2. **Beam-Broadening Techniques for High-Gain Reflector Antennas**, E. W. Smith, J. C. Brand and F. Q. Herschelmann
3. **Phase Error Analysis in Communications Antennas**, P. E. Butzien
4. **Multiple Scattering Contributions of Multiple Circular Conducting Cylinders**

to Antenna Gain Loss, R. L. Moore and B. J. Cown

5. **Guided Waves and Their Consequences for Radome Analysis and Performance**, G. Tricoles, R. A. Hayward and E. L. Rope
6. **Rigorous Asymptotic Analysis of Transmission through a Curved Dielectric Slab**, P. D. Einziger, L. B. Felsen and A. Hessel
7. **Analysis and Applications of Active Microwave Reflectors**, M. Hamid and R. Antebi
8. **A Novel Microstrip Antenna Feed for a Paraboloidal Reflector with Simultaneous RCP and LCP Polarization**, C. C. Post, and K. R. Carver

**URSI SESSION H-2
WAVES IN PLASMAS**

Chairman: Dr. R. W. Fredricks
Commission H

1. **Impedance of a Plane Capacitor in a Beam-Plasma System**, T. Fujita and S. Adachi
2. **Some Amplification Properties of Electromagnetic mm-Waves in a Plasma Tube**, A. Rosenberg, J. Politch, Y. Ben-Aryeh and J. Felsterner
3. **Pulse Propagation in Dispersive Media Brillouin's Signal Velocity**, R. J. Vidmar
4. **A Theory of Coupling Propagation for Electromagnetic and Electron Acoustic Wave in Inhomogeneous Magneto-Plasma**, C. Man and Z. Yan-Zheng
5. **Frequency Shift of a Signal Propagating in a Time-Dependent Environment**, H. Gelman and I. Kohlberg

FRIDAY MORNING, JUNE 19, 1981

**URSI SESSION B-11
ANTENNAS II: ARRAYS AND RADIATING ELEMENTS**

Chairman: Prof. C. M. Butler

Commission B

1. **Analysis of a Slot Radiator in a Dielectric-Covered Parallel-Plate Waveguide**, R. D. Nevels

2. **Radiation Characteristics of Slot Antenna on Large Body-of-Revolution**, T-K Wu
3. **Effect of Array Size and Shape on Element Patterns**, W. K. Kahn
4. **Sidelobe Control in Partially Excited Cylindrical Arrays**, K. J. Keeping and J-C. Sureau
5. **Plane Wave Spectrum Analysis of**

- Coupling Between a Pair of Near-Field Directive Antennas**, B. J. Cown and D. H. Ryan, Jr., and P. A. Major
6. **Mutual Impedance from Far Field Data**, S. Chaiken, G. Franceschetti and N. G. Alexopoulos
 7. **Monopole Antennas on Lossy Ground Planes**, J. D. Lilly and C. A. Balanis

(continued on page 62)

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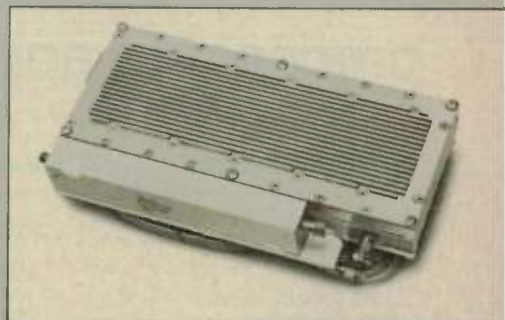
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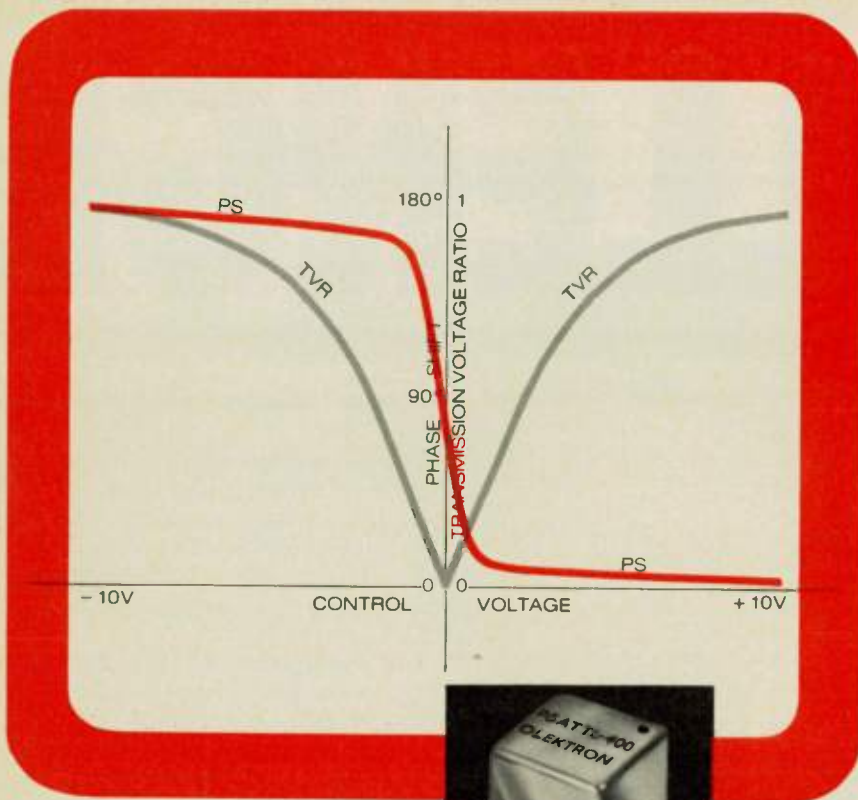
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(from page 60) TECHNICAL PROGRAM

8. Resonance and Q Properties of Isosceles Triangular Patch Antennas of 60° and 90° Vertex, E. P. Kuester and D. C. Chang

9. Microstrip Dipoles on Spherical Structures, N. G. Alexopoulos, and P. L. E. Uslenghi

AP-S SESSION 21
RADOMES/SCATTERING
Chairman: Dr. A. C. Ludwig

1. Measurement, Calculation, and Reduction of Radome Wave Aberrations, G. Tricoles, R. A. Hayward and E. L. Rope
2. The Internal Fields of a Layered Radome Excited by a Plane Wave, Terry M. Kvam, and K. K. Mei
3. Ray Analysis of a Cylindrical Dielectric Radome, P. D. Einziger and L. B. Felsen
4. Scattering of a Surface Wave Incident on a Nonuniform Transition in Surface Reactance, K. S. Park and R. J. King
5. A New Approach to Three-Dimensional Multistatic Target Imaging, Gregory E. Heath
6. Electromagnetic Scattering from a Thin Rectangular Plate, Joseph R. Mautz and Arlon T. Adams
7. Ray Analysis of EM Backscatter from a Jet Intake Configuration, P. H. Pathak and C. C. Huang
8. Efficient Calculation of Scattering from a Hyperboloidal Subreflector, Ronald J. Pogorzelski

AP-S SESSION 22-A
INTEGRAL EQUATION MODELING
Chairman: Prof. Tapan Sarkar

1. Computer Graphics Applications in Electromagnetic Computer Modeling, E. K. Miller, F. J. Deadrick and G. J. Burke, and J. A. Landt
2. A Numerical Analysis of Mutual Coupling Between Thin Linear Antennas in the Presence of a Metallic Scatterer, M. F. Catedra
3. Radiation from Dipoles Near a Finite Length Cylinder, M. A. Hassan
4. Numerical and Experimental Results for Two Thick Monopole Antennas, A. Hizal and M. Qureshi

AP-S SESSION 22 B
LAYERED MEDIA
Chairman: Prof. Tapan Sarkar

1. Plane Wave Scattering from an Infinite Loaded Dipole Planar Array Over a Lossy Half Space, Takunori Mashiko and Saburo Adachi

(continued on page 64)

It's Clear

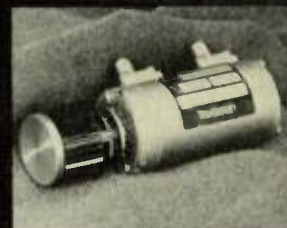
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2. Scattering of EM Waves by Buried or Partly Buried Body of Revolution, Henry Chang, and Kenneth K. Mei
3. Exact Scattering from a Thick Imperfectly Conducting Half-Plane, V. Daniele, and P. L. E. Uslenghi
4. Spatial Fourier Transforms and Equivalent Circuits. A Novel View in Electromagnetic Waves, C. D. Papageorgiou, J. D. Kanellopoulos and J. G. Fikioris

AP-S SESSION 23

WIRE ANTENNAS

Chairman: Dr. A. W. Love

1. An Investigation of the Effects of Ice on Microwave Dipole Receiving Antenna Elements, I. M. Burdeaux, and W. L. Wilson, Jr.
2. Design Method for Fine Impedance Matching of Superturnstile Antenna and Characteristics of the Modified Batwing Antenna, Genpei Sato and Haruo Kawakami, Hirofumi Sato, and R. Wayne Masters
3. Moment Method for a Twin Loop Antenna, Keiji Endo; Genpei Sato, Haruo Kawakami and Mitsuhiro Sato
4. Axial Ratios of Spiral Antennas, Hisamatsu Nakano and Junji Yamauchi
5. Backfire Bifilar Helical Antenna with Tapered Feed End, Junji Yamauchi, Hisamatsu Nakano and Hiroaki Mimaki
6. Unipole-Notch Array Antennas, Takayuki Ishizone, Katsuhiko Yamamoto, Kunio Sawaya and Yasuto Mushiaki

7. Frequency Response of Uniform and Non-uniform Helical Antennas Radiating in the Axial Mode, Ping-Fai Wong, Kang-Fong Lam and Kai-Fong Lee
8. Driving-Point Impedance of Corner-Reflector Antenna with Tilted Dipole, Kai-Fong Lee

AP-S SESSION 24

REFLECTOR ANTENNAS — II

Chairman: Dr. James W. Duncan

1. Optimum Feed Locus for General Three-Dimensional Beam Scanning in Offset Cassegrain Antenna, V. Krichevsky and D. F. DiFonzo
2. On the Computation of the Radiation Pattern of Shaped Cassegrainian Reflector Systems, R. Mittra, W. L. Ko and M. S. Sheshadri, and V. Galindo-Israel
3. The Design of Wide-Band, Sharp Cut-Off Dichroic, C. A. Chen, P. G. Ingerson and C. C. Chen
4. Millimeter-Wave Beam Steering by Motion of Cassegrain Antenna Subreflector, R. W. McMillan, R. C. Rogers, III and G. F. Smith
5. A Scanning Offset-Fed, Near-Field Gregorian Reflector Antenna, Yeongming Hwang and C. C. Han
6. A Study of Gravitational Astigmatism Compensation of Large Cassegrain Antennas, Shun-Shi Zhong
7. Influence of the Beam Waveguide System on the Cross-Polar Radiation from Symmetrical Reflector Antennas, Ynag Kezhong

URSI SESSION B-12
MEDIA AND EARTH INTERACTION

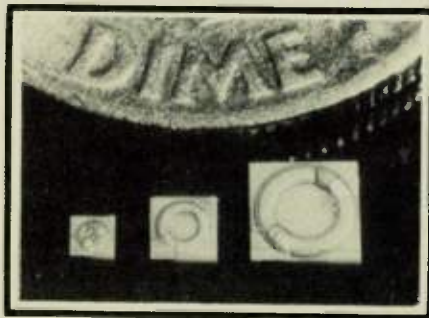
Chairman: Dr. G. Brown

Commission B

1. Transmitter and Receiver Aperture Averaging Effects for the Intensity Fluctuations of a Beam Wave in the Turbulent Atmosphere, Y. Baykal, C. F. Ouyang, M. A. Plonus and S-J Wang
2. A Coordinate-Free Approach to Wave Reflection from a Moving Medium, H. C. Chen
3. Scattering from a Random Slab, H. M. Huby and K. Sivaprasad, and R. Vasudevan
4. Attenuation Constant of Coherent Field in Dense Spherical Particles, Y. Kuga and A. Ishimaru
5. Scattering Cross Sections for Composite Random Surfaces — Full Wave Analysis, E. Bahar
6. Bistatic Scattering from a Lossy, Random, Composite Surface, G. S. Brown
7. An Improved Model for Scattering from Rough Terrain, R. J. Papa, J. F. Lennon and R. L. Taylor
8. A Dual Approach for Wave Packets and Solitary Waves in Nonlinear Systems, D. Censor
9. Magnetic Field Analysis of an Infinite Line Current Over a Plane Layered Earth, O. Aboul-Atta, L. Shafai and M. Z. Tarnaweky

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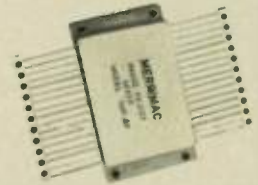
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 DC to 18 GHz



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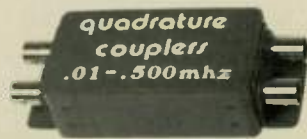
0/180 Hybrid Tee Junction
 2 MHz to 18 GHz



Directional Couplers
 100 kHz to 18 GHz



Quadrature 90 Couplers
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 (3 dB Hybrid Dividers/
 Combiners)

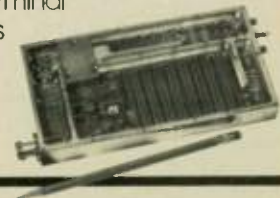


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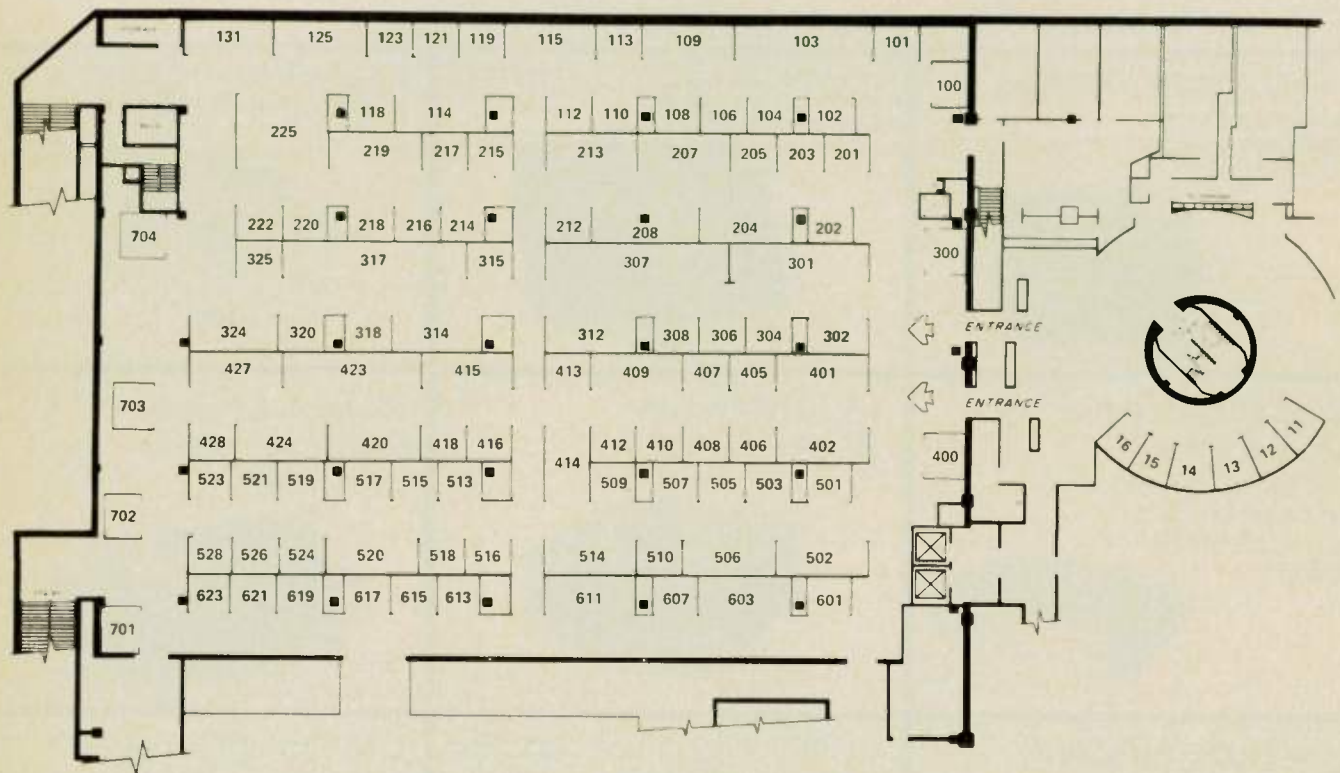
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World Radio History
 CIRCLE 54 ON READER SERVICE CARD

EXHIBITION



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A.E.G. Telefunken Corp.	121	Compact Engineering, Inc.	324	Integra Microwave	213
Aercom Industries	307	Compex Corp.	601	International Microwave Corp.	216
Airtron Div. Litton Industries	208	Daden Associates	519	Isotronics	109
Alford Manufacturing Co.	307	Delta Microwave	306	Johanson Mfg. Co.	400
Alpha Industries, Inc.	424	Dexcel, Inc.	108	Journal of Electronic Defense	207
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American Technical Ceramics	214	Dielectric Labs., Inc.	413	K & L Microwave, Inc.	215
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Amplicon Inc.	621	EEV, Inc.	516	Keene Corp., Ray Proof Div.	613
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Anzac Electronics/Adams Russell	302	Electronic Navigation Industries	519	Litchfield Microwave Lab	12
Applied Engineering Products	702	Engelmann Microwave Co.	106	Locus, Inc.	603
ARRA Inc.	307	EPSCO, Microwave Inc.	406	Logimetrics, Inc.	304
Artech House	207	Epsilon Lambda, Electronics Corp.	325	Lorch Electronics	15
Automatic Connector, Inc.	100	EPSCO, Microwave Inc.	406	Luxtron	703
Avantek, Inc.	420	Filtronics	509	3M Company	204
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Cablewave Systems, Inc.	517	W. L. Gore & Associates, Inc.	125	Mauzy Microwave Corp.	115
California Eastern Labs, Inc.	312	Hewlett-Packard Co.	317	McDougall/McLaughlin Co.	519
Central Microwave	519	Honeywell, Inc.	515	Merrimac Industries, Inc.	405
Cleland & Co.	603	Horizon House	207	Metelics	15
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		Hughes Aircraft Co., El. Dyn. Div.	409	Micronetics, Inc.	307
				Micro-Now Instrument Co.	524

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Microwave Exhibitions & Publishers Ltd.	617
Microwave Journal	207
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Div. Cleland Co.	
Microwave Systems News	607
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Millis Research	307
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Mitsubishi Electronics	112
Narda Microwave Corp.	502
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Pacific Measurements Inc.	102
Parametric Industries, Inc.	104
Passive Microwave Technology	519
Plessey Optoelectronics & Microwave	528
Polarad Electronics, Inc.	513
Premier Microwave	307
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Raytheon Co.	314
RF Associates, Inc.	307
RHG Electronics Lab Inc.	15
RL Components, Inc.	701
Rogers Corp.	113
Sage Labs, Inc.	501
Sivers Lab	603
Solid State Technology	307
Soliton/Microwave	208
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Syston-Donner Corp., Adv. Comp. Div.	415
Syston-Donner Corp. Microwave Div.	704
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Tekform Products Co.	222
Tektronix Inc.	203
Telecommunications	207
Teledyne Microwave	205
Texscan Corp.	307
Thinco, Div. Hull Corp.	601
Thomson CSF Comp. Corp. Electron Tube Div.	423
Thomson CSF Comp. Corp. Semiconductor Div.	423
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TRAK Microwave Corp.	505
TRW RF Semiconductors	427
Tranco Products Inc.	510
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J-Z, Div. Dynatech	416
Uniform Tubes	318
Varian Associates, EDG	225
Vectronics Microwave Corp.	16
Vichem Corp., Gel-Pak Div.	123
Natkins-Johnson Co.	506
Navetek	103
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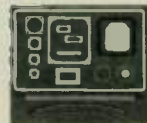
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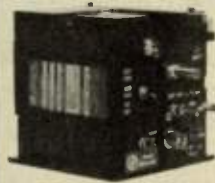
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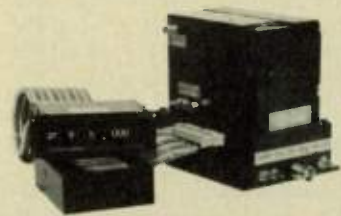
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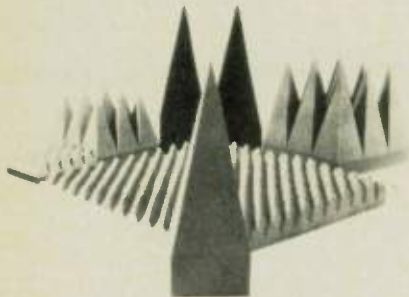
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EXHIBITION GUIDE

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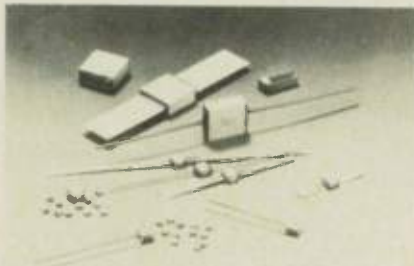
AMERICAN ELECTRONIC LABORATORIES, INC. Lansdale, PA

T. Keffer, L. Fisher, J. Ciral, G. Seiver, M. Nussbaum, L. McCreddin, D. Martinson, T. Clark

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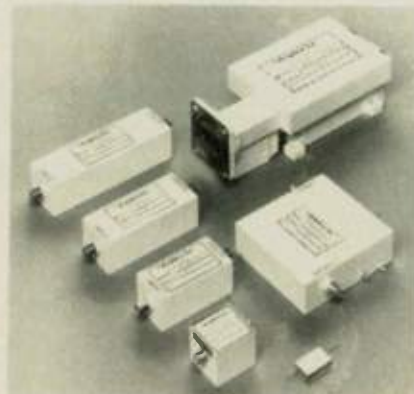
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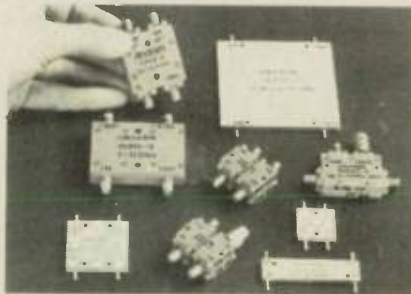
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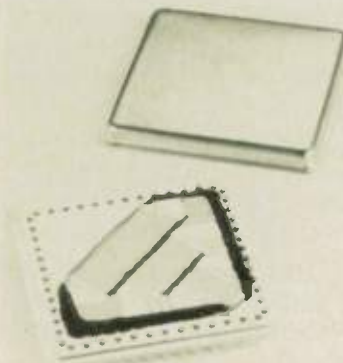
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(continued on page 72)

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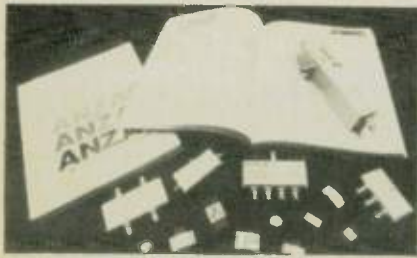


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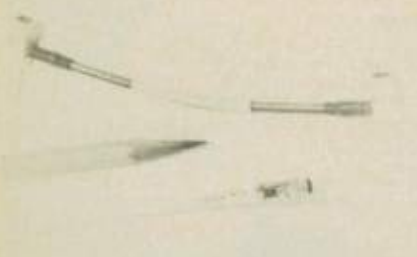
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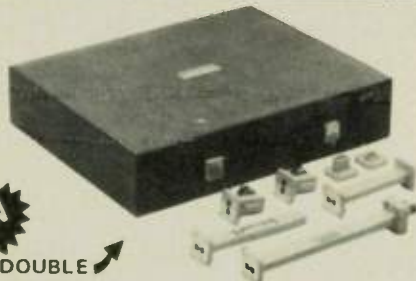
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(continued on page 74)

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2450B	2450K	GR900	DC-8.5
2650B	2650K	APC7	DC-18
8050A	8050B	APC3.5	DC-18
8150B	8150K	MPC8(OSSM)	DC-18
8440B	8440K	HN	DC-8
8450B	8450K	SC	DC-10
8490B	8490K	C	DC-10
8550B	8550K	BNC	DC-12.4
8650C	8650K	TNC	DC-18
8750C	8750K	SMA	DC-18
8850B	8850K	N	DC-18

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Kit Model Number	WR Desig.	Freq. Range GHz
R7005	430	1.7 - 2.6
S7005	284	2.6 - 3.95
E7005	229	3.3 - 4.9
G7005	187	3.95 - 5.85
F7005	159	4.9 - 7.05
C7005	137	5.85 - 8.2
H7005	112	7.05 - 10.0
X7005	90	8.2 - 12.4
M7005	75	10.0 - 15.0
P7005	62	12.4 - 18.0

DOUBLE RIDGED W/G KITS

MODEL NUMBER	WAVEGUIDE DESIGNATION	FREQ RANGE GHz
DA 7005	WRD 750	7.5 - 15
DB 7005	WRD 475	4.75 - 11
DC 7005	WRD 350	3.5 - 8.2

MM W/G KITS

MODEL NUMBER	WAVEGUIDE DESIGNATION	FREQ RANGE GHz
A7005	WR42	18 - 26.5
U7005	WR28	26.5 - 40



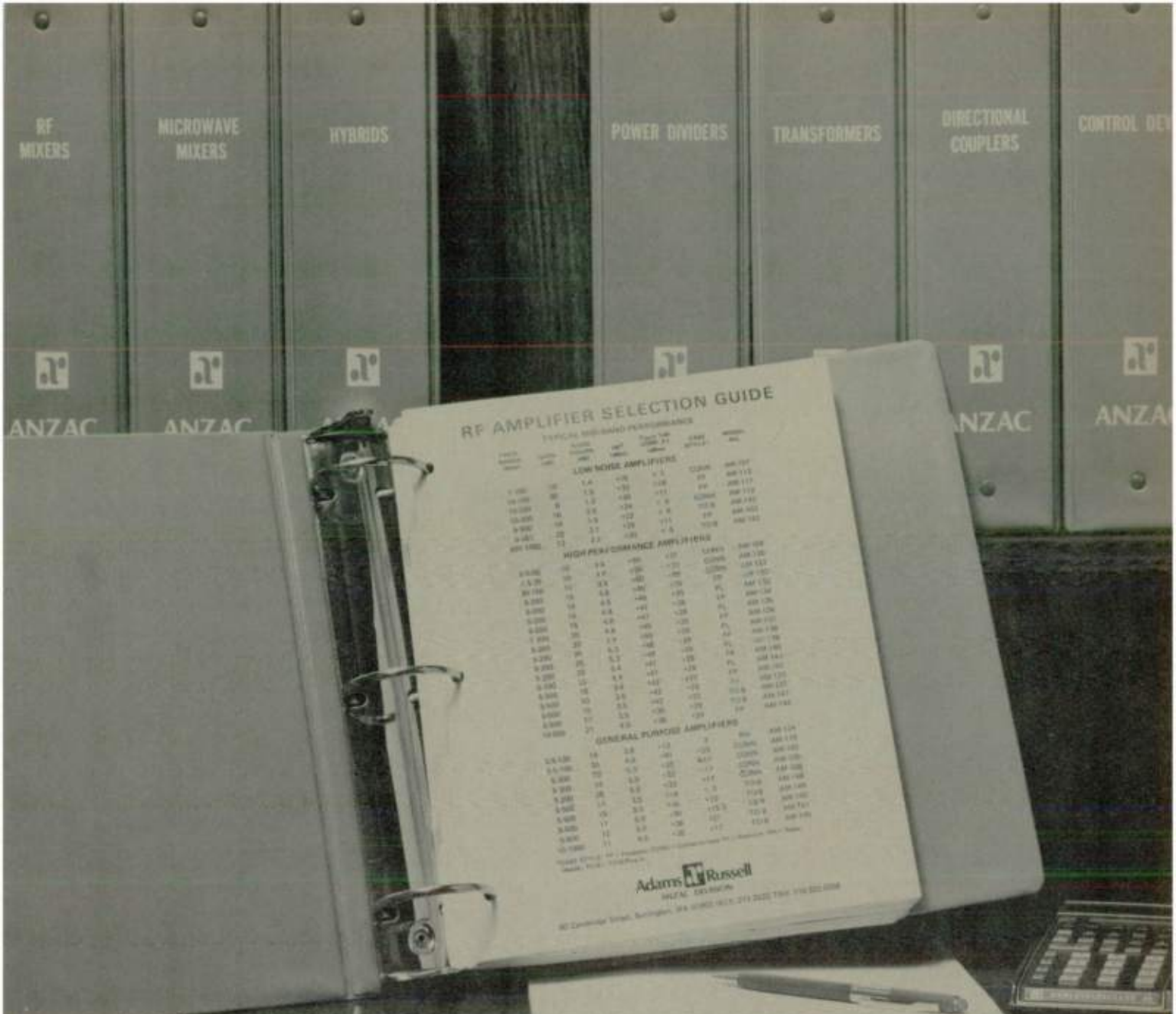
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BAYTRON COMPANY, INC. 218
Medford, MA
R. McLaughlin, M. F. Kozul, M. Kozul, T. Kozul



Millimeter devices.

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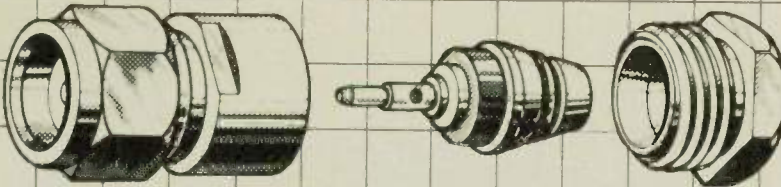
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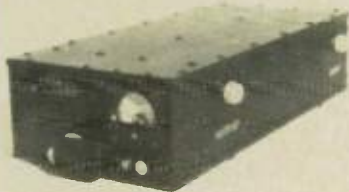
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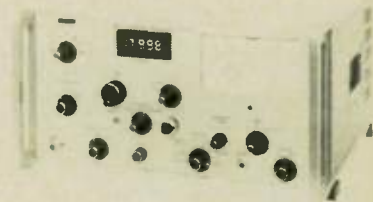
EATON CORP.
Electronic Instrumentation Div.
Los Angeles, CA

J. Molt, W. Purdy, C. Sorensen

DIELECTRIC LABS INC. 413
Cazenovia, NY

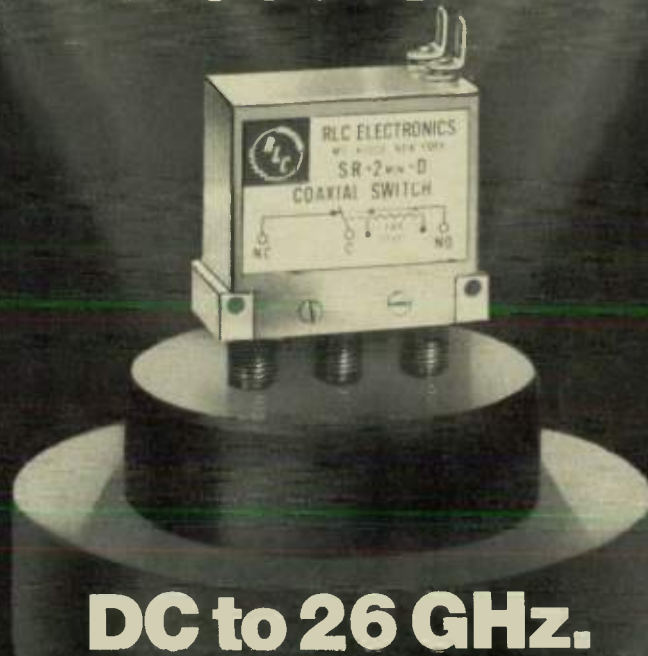
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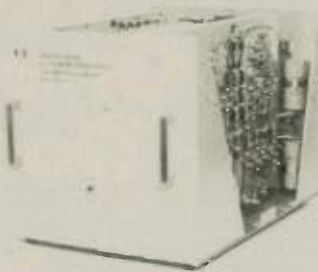
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(continued on page 78)

... and other components for your system



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(from page 77) GUIDE

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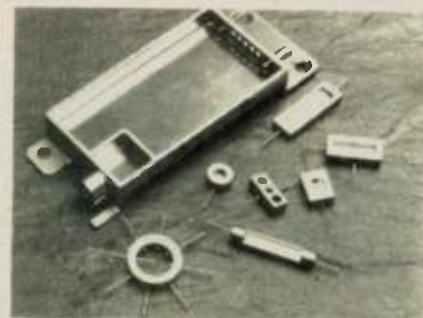
L. Johnson, W. Schuerch

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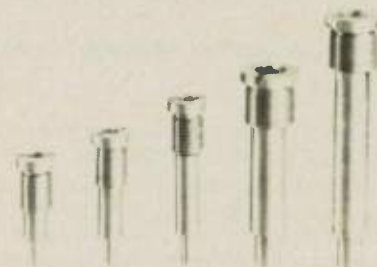
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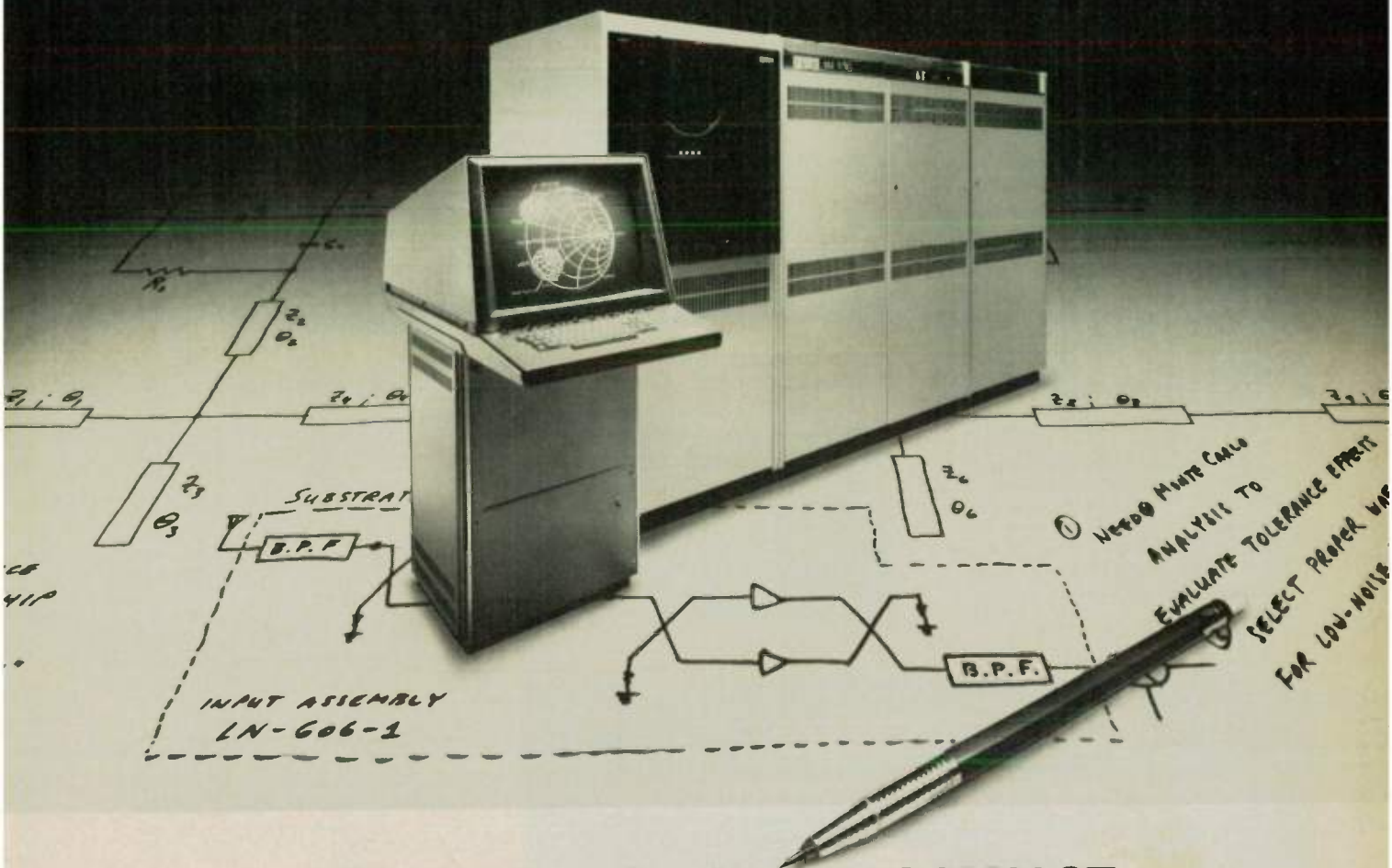


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12

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Detectors, limiters, diode switches (less drivers), limiter-detectors, switch-limiter-detectors, isolators, comb generators, coupler detectors, low frequency VCO's to 2 GHz.

613

T. Russell, C. Gentile, M. Molley, D. Yoshii



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KW ENGINEERING, INC.
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319

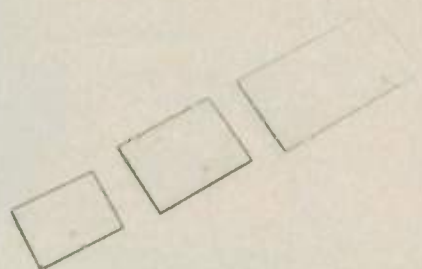
A. Brand, P. Kovacich, K. Wittig, D. Clark



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603

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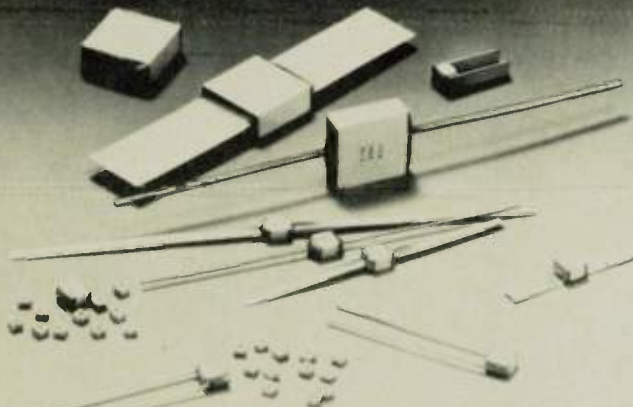
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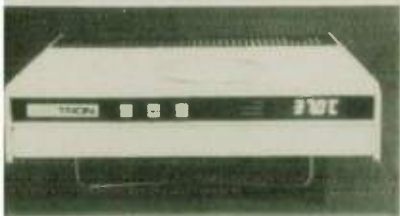
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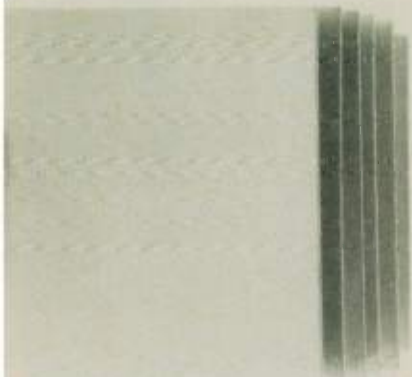
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301

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(continued on page 82)

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TO-5 package. Single and double-balanced. Three LO drive levels: +3 dbm, +7 dbm, +17 dbm. RFI shielded. Hermetically sealed. Frequencies up to 1,500 MHz.

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ENI

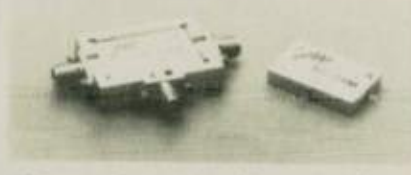


The advanced
design line of
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(from page 81) GUIDE

MAGNUM MICROWAVE CORP. 118
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D. Fealkoff, H. McGrath, H. Soza



Microwave signal processing components, mixers, oscillators, amplifiers.

MAST MICROWAVE 217
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Complete line rotary joints from dc to 40 GHz, single and multichannel, waveguide and coaxial type transmission line.

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Los Angeles, CA

Merle McDougall, Dick McLaughlin.

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MERRIMAC INDUSTRIES, INC. 405
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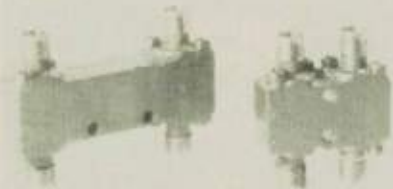
M. Asnes, A. Egger, G. Tahlmore



This comprehensive 320 page catalog describes our complete line of IF and RF frequency components and subassemblies using distributed element and lumped element designs.

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Norwood, NJ

G. Simonyan, L. Fisher, J. Ciral, G. Seiver, D. Martinson, L. McCredlin

(continued on page 86)

THERE ARE STILL SOME RADAR SYSTEMS WE DON'T MAKE A RADOME FOR.



Increasingly, designers and engineers are including protective radomes in their plans for new radar systems. With good reason.

They've found that radome-enclosed radar systems provide better performance over considerably longer periods of time than exposed radar systems.



Radomes dramatically reduce radar failure rates, prolong system life, and allow

routine maintenance without weather interruptions.

And radomes prove to be a cost-effective investment that pays for itself quickly. Radar system maintenance and repair costs are cut considerably. Severe damage from heavy snow, strong winds, and massive ice build-ups is eliminated. As are the corrosive effects of sand, salt, dust, and chemical pollutants.

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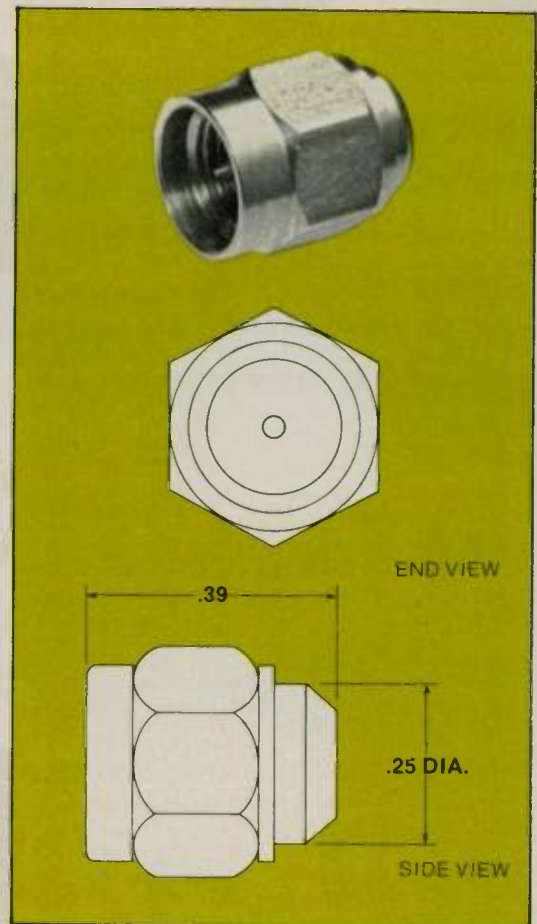


SMA TERMINATIONS

0.5, 1, 2, 5 AND 10 WATTS

- DC to 18.0 GHz
- SMA Connectors
- 1.05 + 0.008 fGHz Maximum VSWR
- -54°C to +125°C
- Excellent Reliability
- Stainless Steel Connectors

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- Connector: SMA M/F
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- Construction:
 - Connector: Stainless Steel
 - Body: Stainless Steel
- Length, Maximum: 0.44"
- Diameter: 0.25"
- Delivery: Stock to 30 days ARO



*ALL MMI SERIES TERMINATIONS ARE AVAILABLE WITH 3.5" BEAD CHAINS

-add suffix "C" to model number

** Add M or F to model number to specify Male or Female.



MMI MODEL 2055* **

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- Frequency Range: DC to 18.0 GHz
- Connector: SMA M/F
- Impedance: 50 ohms
- VSWR, Maximum: 1.05 + 0.008 f(GHz)
- Temperature Range: -54°C to +125°C
- Construction:
 - Connector: Stainless Steel
 - Body: Stainless Steel
- Length, Maximum: 0.58"
- Diameter, Maximum: 0.29"
- Delivery: Stock to 30 days ARO



MMI MODEL 2013* **

- Input Power @ 25°C: 10.0 watts
- Frequency Range: DC to 18.0 GHz
- Connector: SMA M/F
- Impedance: 50 ohms
- VSWR, Maximum: 1.05 + 0.01 f(GHz)
- Temperature Range: -54° to +125°C
- Construction:
 - Connector: Stainless Steel
 - Body: Black Anodized Aluminum
- Length, Maximum: 1.62"
- Diameter, Maximum: 1.30"
- Delivery: Stock to 30 days ARO



MMI MODEL 2057* **

- Input Power @ 25°C: 2.0 watts
- Frequency Range: DC to 18.0 GHz
- Connector: SMA M/F
- Impedance: 50 ohms
- VSWR, Maximum: 1.05 + 0.008 f(GHz)
- Temperature Range: -54° to +125°C
- Construction:
 - Connector: Stainless Steel
 - Body: Black Anodized Aluminum
- Length, Maximum: 0.92"
- Diameter, Maximum: 0.50"
- Delivery: Stock to 30 days ARO



MMI MODEL 2010* **

- Input Power @ 25°C: 5.0 watts
- Frequency Range: DC to 18.0 GHz
- Connector: SMA M/F
- Impedance: 50 ohms
- VSWR, Maximum: 1.05 + 0.01 f(GHz)
- Temperature Range: -54° to +125°C
- Construction:
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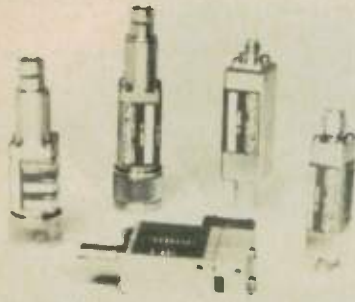
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01-836-1228

ISRAEL: Racom Electronics Co., Ltd.
03-453151

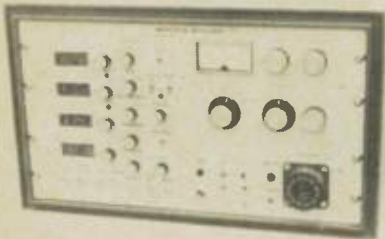
JAPAN: Toko Trading, Inc.
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Solid state noise sources — diodes, generators and programmable noise sources; Microwave Components — WG and coax switches; Dummy loads, crystal detectors; bolometers.

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Miniature band pass, low pass and high pass filters, contiguous multiplexers, (both comb line and stub types) back detectors, Schottky detectors, video detector logarithmic amplifiers, threshold detector amplifiers, microwave integrated assemblies and 18 to 40 GHz components.

MICROTEK CO. LTD. 13
Bangkok, Thailand
Microwave ferrite isolators and circulators.

MICROWAVE APPLICATIONS GROUP 308
Santa Maria, CA
C. Boyd, W. R. Reed, T. Stokes



Rotary-field phase shifter and dual-mode phase shifter line.

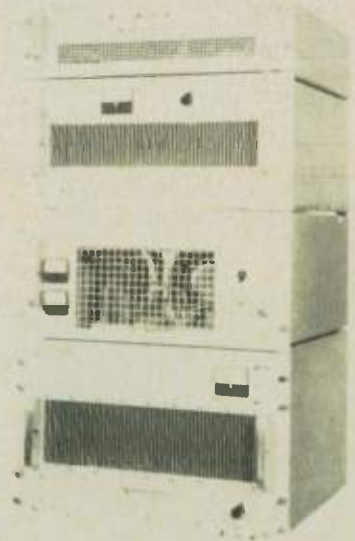
MICROWAVE EXHIBITIONS & PUBLISHERS LTD. 617
Sevenoaks, Kent, UK
R. Marriott

International exhibition and conference organizers for the microwave community.

MICROWAVE JOURNAL 207
(see Horizon House)
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W. Bazy, C. Berman, B. Bossard, J. Cotsworth, H. Ellowitz, C. Hagen, R. Howland, E. Johnson, S. Pasqualucci
Microwave Journal Magazine, subscription and editorial information.

MICROWAVE POWER DEVICES, INC. 300
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D. Mazziota, W. Liebman, R. Sheloff



Solid state RF and microwave power amplifiers and related subsystems for military space and industrial applications.

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High-power,
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modulator
Model 2076



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- High-power, pulsed cavity oscillator with MIL-qualified solid-state modulator
- Requires only a TTL-compatible input pulse and various supply voltages
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- Single knob tuning, four-digit readout

High-power,
pulsed
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Model EP5000L



For use in:

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- Delivers up to 5000 W of power
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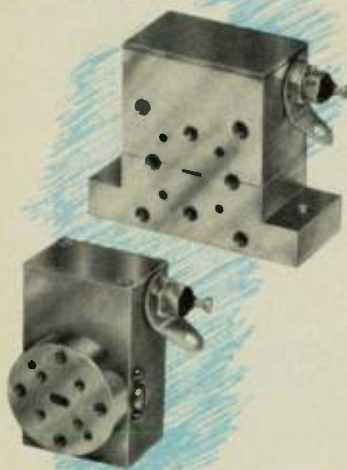
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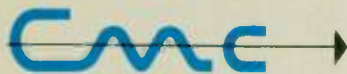
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(from page 86) GUIDE

PIN and Schottky diodes and diode modules.

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Los Angeles, CA

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Manufactures and designs simple waveguide assemblies, machined parts, microwave devices and electro/mechanical assemblies.

MICROWAVE SYSTEMS NEWS 607
Palo Alto, CA

J. Collins, B. Eustace, H. Eustace, J. Fawcette, C. Flinn, H. Martin, M. Smith, A. St. John, Mr. and Mrs. A. Yaconetti
Magazine.

MICROWAVES 114
Rochelle Park, NJ

S. Bearse, H. Bierman, W. Bojsza

MicroWaves Magazine, MicroWaves Product Data Directory.

MIDISCO 526
Commack, NY

A. L. Leiman, D. Leiman

RF components, including attenuators, cable assemblies, coaxial detectors, shorts, terminations, directional couplers, hybrids, mixers, phase adjusters, RFI shielded cases, and inter/intra-series adapters.

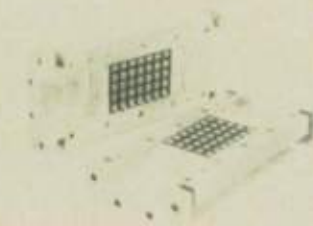
MILLIS RESEARCH 307
Millis, MA

J. Ciral, L. Fisher, D. Martinson, L. McCreddin, L. Searle, G. Seiver

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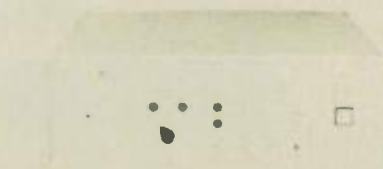
M. Bentob, T. Robinson



Waveguide, coaxial components and sub-systems, transmission line systems, complete antenna systems for terrestrial microwave systems and ground satellite stations, and microwave instrumentation.

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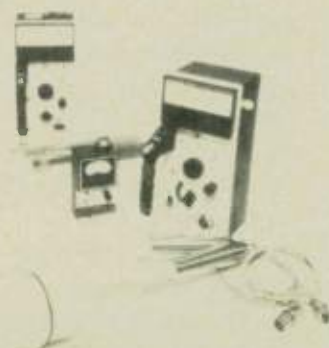
MOTOROLA, INC. 14
Phoenix, AZ

Alan Wagstaffe, Steve Lazar, Ed Reimanis, Tom Bishop

RF microwave transistors.

THE NARDA MICROWAVE CORPORATION 502
Plainview, NY

J. Coppola, W. Bourke, A. LeMay, G. DeBella, G. Gianasio, C. Buntschuh, E. McSweeney, J. Reid, W. Lamb, R. Vaccarino



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(continued on page 90)

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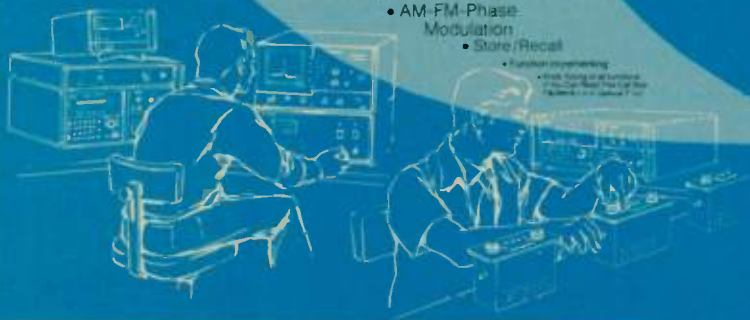
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OMNI-WAVE ELECTRONICS 307
Gloucester, MA

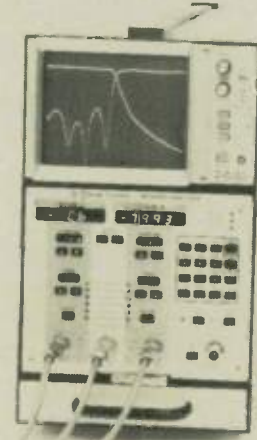
J. Ciral, M. Hamilton, L. Fisher,

D. Martinson, J. Fossiano, L. McCreddin,
G. Seiver, S. Brennan

Microwave radar components, duplexers,
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PACIFIC MEASUREMENTS INC. 102
Sunnyvale, CA

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PASSIVE MICROWAVE 519
TECHNOLOGY
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George Grund, Bob Weiss

Coaxial and waveguide isolators and
circulators.

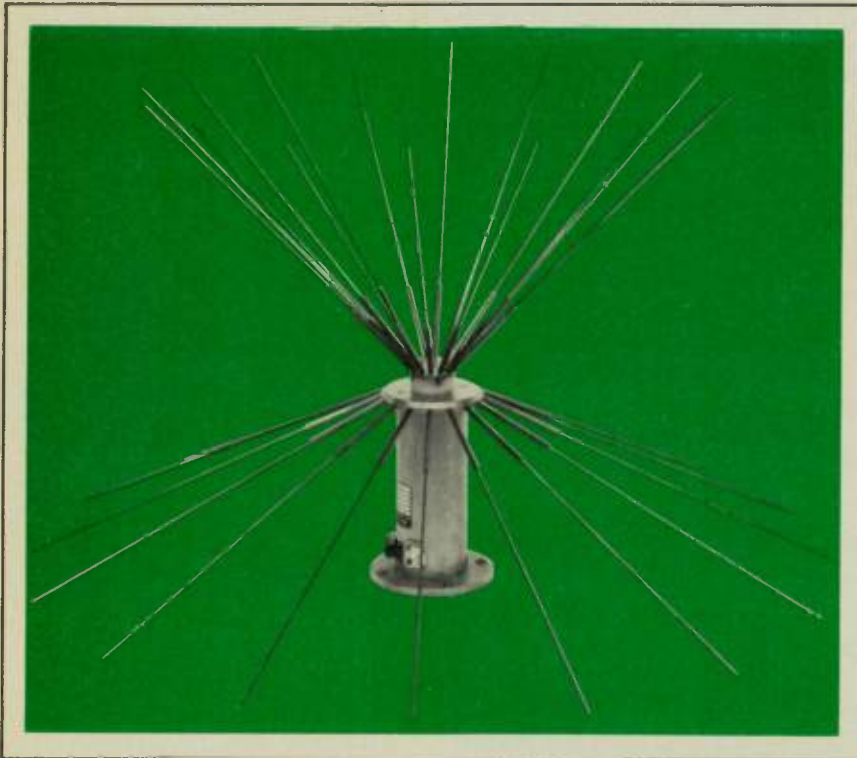
PLESSEY OPTOELECTRONICS 528
AND MICROWAVE
Irvine, CA

P. Minett, E. Bell, D. Bell, G. Youngdahl



(continued on page 91)

MICROWAVE JOURNAL



20Hz to 1GHz: 1 small antenna

ELECTRICALLY SMALL ANTENNAS: NEW FROM TECOM

Electrically small antennas whose elements are an extremely small fraction of λ . Among other things, they give you low frequency performance in a small package. Type 201191, shown, has 18 inch removable elements, weighs only 20 pounds, and stows in a convenient carry case. For SIGINT surveillance, it's a vertically polarized, biconical omnidirectional antenna that provides 2 outputs: 20 Hz to 100MHz, and 70MHz to 1 GHz. The Type 201191 antenna — available with optional amplifier and radome.

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more announcements and call or
write for more information.

TECOM

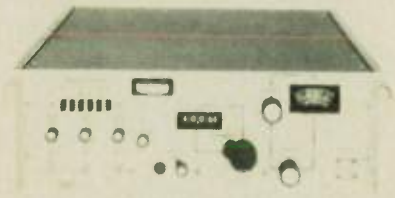
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M. Rogers, D. Harrey, G. Callaway, H. Mead



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Microwave tubes, solid state microwave products, ferrite devices.

RF ASSOCIATES, INC. 307

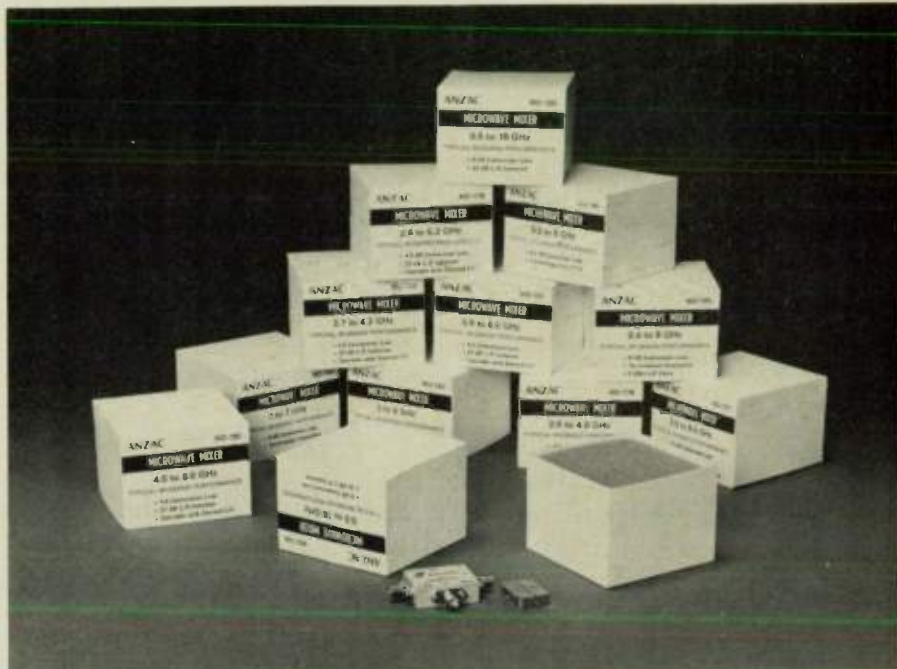
Los Angeles, CA

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Deer Park, NY

R. Hirsch (continued on page 92)



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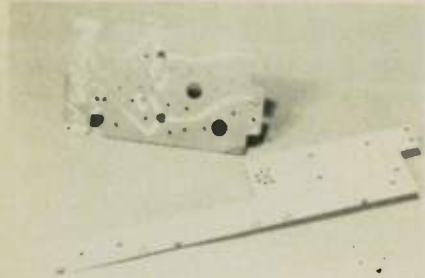
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Marina del Rey, CA

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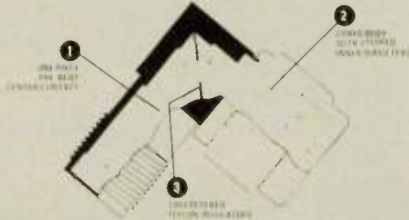
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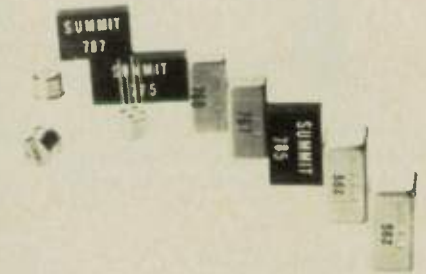
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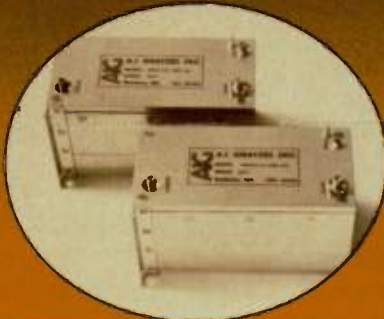
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(continued on page 94)

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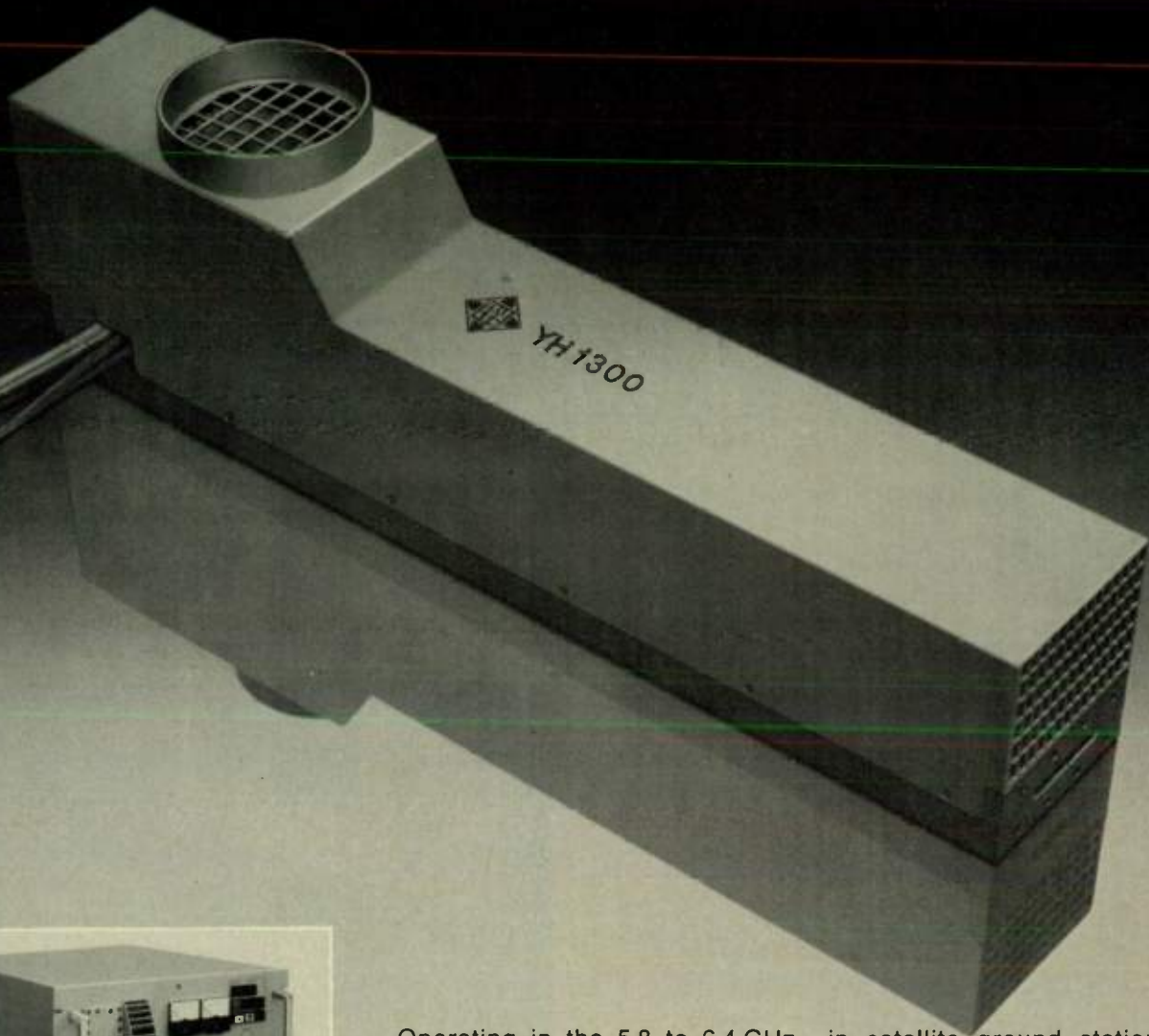
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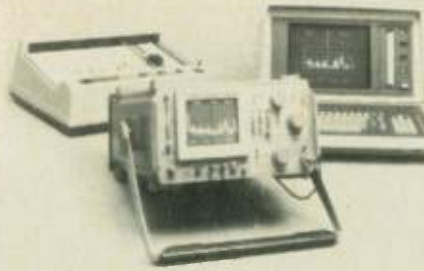
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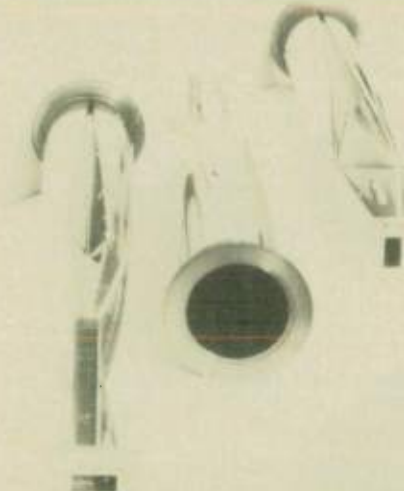
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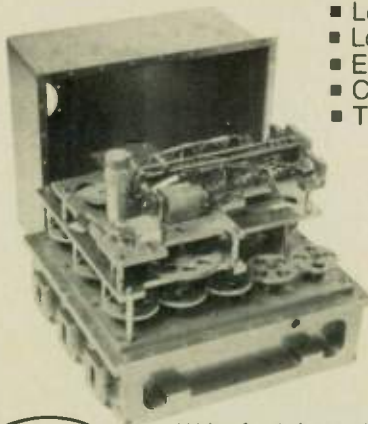
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(continued on page 96)

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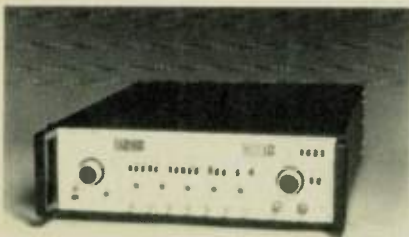
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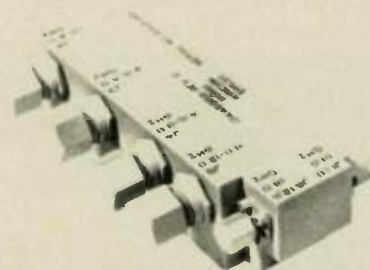
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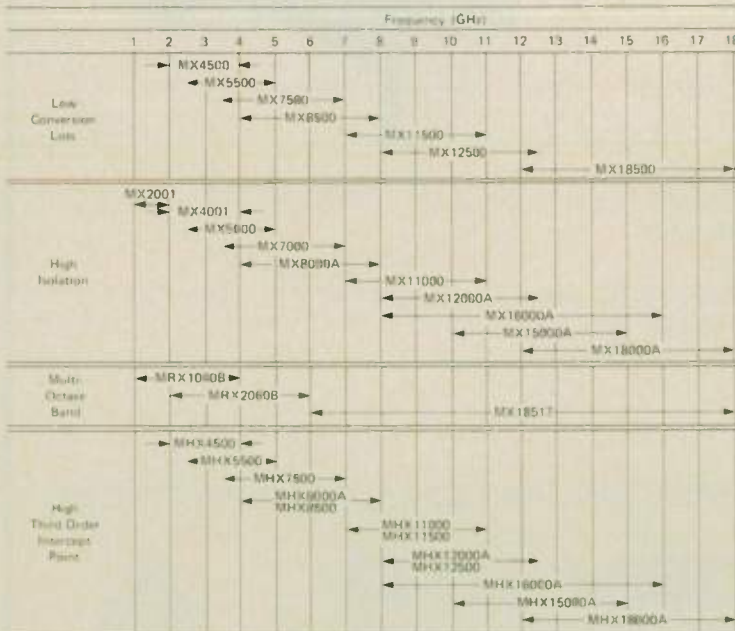


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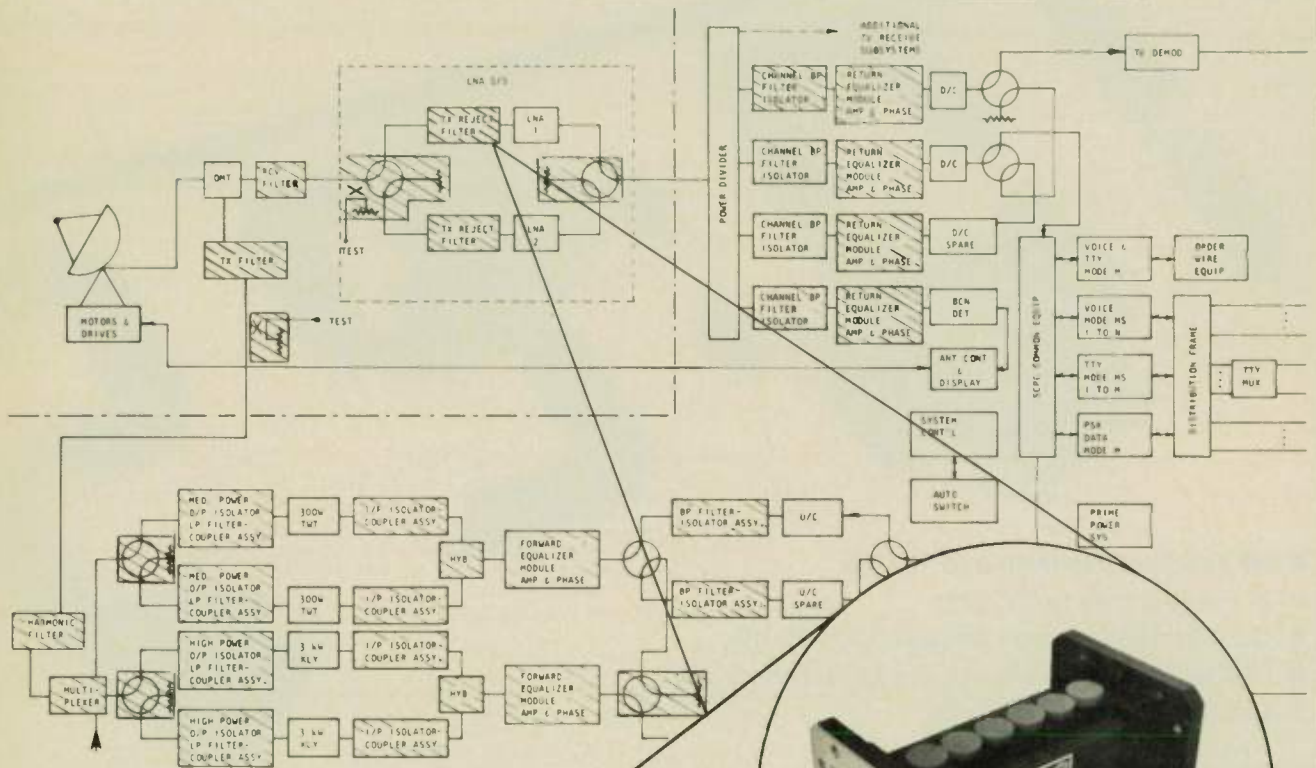
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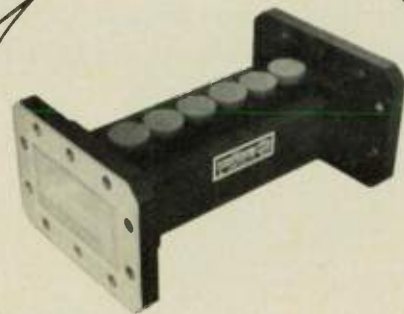
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GERALD GREEN, Washington Editor

DARPA Delineates Advanced Technologies in FY 1982 Budget Request

Since the Defense Advanced Research Projects Agency (DARPA) serves in a "corporate research" role for the Department of Defense (DoD), the agency's projects provide good guidance as to the direction of future technology applied to military systems.

This assessment is certainly borne out when one considers the presentation to Congress by Dr. Robert R. Fossum, the Director of DARPA. In making his presentation to the Research & Development Subcommittee of the House Armed Services Committee, in support of DARPA's fiscal year 1982 budget request, Dr. Fossum delineated a roadmap of future "revolutionary" technology being studied and developed for military applications. Of particular interest to the microwave industry:

- **New Transistors** A new ultra-high frequency transistor, called the permeable base transistor (PBT), is believed capable of operating at frequencies above 500 GHz. A preliminary version of the PBT has been developed and demonstrated to operate at frequencies above 30 GHz.

The smallest feature (0.1 micron) silicon transistor ever fabricated was realized through a highly innovative process technology that uses conventional optical lithographic techniques. The feasibility of using highly focused ion beams to fabricate submicron feature devices without the use of photomasks was demonstrated.

- **Monolithic Gallium Arsenide (GaAs) Integrated Circuits (IC's)** These circuits, based on GaAs rather than silicon, offer substantial speed and/or power advantages if the problems in fabrication that control circuit yield can be resolved. The results of this program could permit space surveillance programs to meet their goals for low-power and high-radiation-tolerant signal processing digital integrated circuits.

The process technology underlying the DARPA program of direct ion implantation into semi-insulating GaAs substrates has been adopted industry-wide and is utilized by R&D programs of all three military services and the intelligence community. Solid-state radar transceiver IC's are also evolving from this research and will affect space-based radar, airborne radar, and electronic countermeasures systems.

- **High Resolution Calibrated Airborne Measurement Program (HI-CAMP)** The HI-CAMP sensor is an advanced, infrared measurement system designed to make high resolution, two-dimensional measurements of earth backgrounds and mobile and stationary air, sea, and land targets. The HI-CAMP sensor has two types of advanced two-dimensional monolithic infrared arrays with charge coupled device multiplexers capable of providing data in two broad infrared bands.

HI-CAMP will demonstrate the infrared charge coupled device (IRCCD) technology and develop a two-dimensional data-base of high spectral, spatial, and temporal resolution background and selected target measurements.

- **Space-Based Radar** Developments have been in progress in such critical areas as transceiver modules, antenna membranes, and target identification. The program is oriented towards component technology and will consider major subsystem development. Achievements to date include the development and testing of a three-layer antenna membrane that is considered a breakthrough in space antenna development. In the area of transceiver development, significant breakthroughs have been achieved in the areas of weight and cost.

The total estimated budget for DARPA in FY 1982 is \$655.0 million. The estimated FY 1983 budget is \$691.3 million.

ECAC Move Studied

The Air Force is studying relocation of the Electromagnetic Compatibility Analysis Center (ECAC) from Annapolis, MD to Duluth, MN, as part of a plan to close its active installation at Duluth International Airport (IAP).

ECAC is DoD's center of expertise on electromagnetic compatibility and plays an important role in an effort to reduce interference between friendly, as well as enemy transmissions. ECAC's role is becoming steadily more important since the number of signals being transmitted is increasing at a rapid rate.

According to the Air Force study, 771 military, civilian, and contractor personnel would be transferred to Duluth, if the planned relocation is approved.

News from Washington

RFPS Issued for SPADOC Four

Request for Proposals (RFP's) for a multi-million dollar contract have been issued to companies interested in becoming contractors for the Space Defense Operations Center - Phase 4 (SPADOC - 4) Program by the Air Force Systems Command's Electronic Systems Division, Hanscom AFB, MA.

The SPADOC - 4 program will consolidate command, control and communications functions from many sources allowing SPADOC personnel to determine actions and issue timely directions for Space Defense Operations.

SPADOC - 4 will be implemented via an "evolutionary acquisition strategy" that is designed to build on the experience gained by actual system-use in its own operational environment.

According to Lt. Col. Peter R. Wilkinson, Program Manager, the program will involve a one-year development segment. During development, specific operational capabilities will be delivered through a series of "acquisition blocks."

A contract award for the definition part of the program is expected to be awarded by November 1981.

Navy Awards \$3.9 Million JTIDS Contract

The Naval Electronics Systems Command has awarded a \$3.9 million contract to TADCOM, a joint venture of ITT and Hughes Aircraft Company, for design studies of several classes of terminals for the Joint Tactical Information Distribution System (JTIDS).

The JTIDS military information exchange system, sponsored by the Department of Defense (DoD), will allow US naval shipboard, ground, and airborne forces to communicate over voice and data channels that are secure and jam-resistant. The system also provides navigation information that will give grid lock and facilitates identification of friendly or hostile forces.

Robert Kramp, President of TADCOM, said the contract involves design of engineering development units of Class 1 surface command terminals, Class 1A airborne command terminals and Class 2 airborne tactical terminals.

The terminals will operate on the principle of both distributed time division multiple access (DTDMA), an ITT development, and time division multiple access (TDMA).

DoD has encouraged formation of industry teams as a means of developing dual, highly qualified sources for volume production of JTIDS terminals.

Development of JTIDS terminals has been performed by the Naval Electronic Systems Command under the direction of Capt. Robert May, USN. TADCOM's Program Director is Frederick A. Modavis.

British Aerospace and COMSAT Gen. to form Joint Venture

Following a Memorandum of Understanding signed several months ago, COMSAT General Corporation (CGC) and British Aerospace (BAe), in anticipation of the award of a contract to provide leased communications services to the British Ministry of Defence, have announced their plans to form a joint venture.

COMSAT General and British Aerospace have a long history of involvement in the field of satellite communications. Over the past 15 years CGC established communications system comprising a total of 32 satellites. BAe has been chosen as prime contractor for all European Space Agency communications programs.

COMSAT General is also currently under contract to the UK Royal Navy to provide defence communications service via satellite through the MARISAT System. Established in 1976, the MARISAT System provides worldwide service to the U.S. Navy and the commercial maritime and offshore industries.

It is considered likely that future defense communications systems spawned by this joint venture could result in a substantial increase in the UK aerospace industry's participation in the design and construction of major satellite programs throughout the world.

USAF Group Formed to Study Atmospheric Effects on Radio and Radar

A newly formed Air Force Joint Management Group on Tropospheric Propagation will focus on natural events which impact on radio and radar communications.

Chaired by Capt. Ruth B. Kaplan of the Electronic Systems Division's Deputy for Development Plans at Hanscom AFB, MA, the group brings together hardware-engineering skills of the Electronic Systems Division, propagation and engineering research knowledge of the Rome Air Development Center, atmospheric research experience of the Air Force Geophysics Laboratory, and field-test environment and operational support of the Air Weather Service. ☛

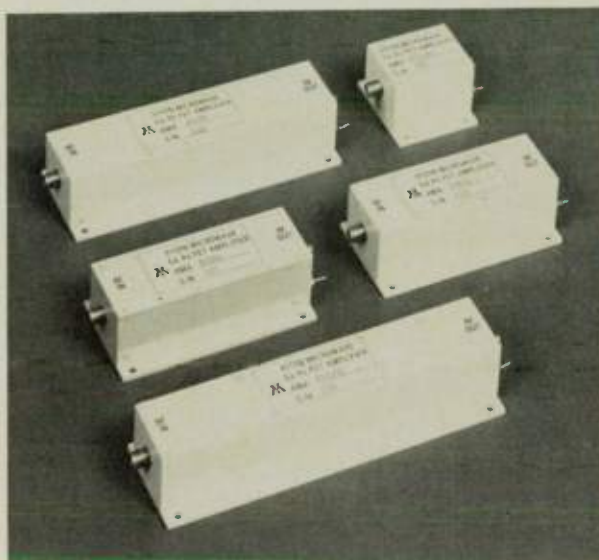
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For more information about Aydin Limiting Amplifiers as well as other Aydin Solid State Amplifiers, contact Hal Sawert at (408) 946-5600, or write:

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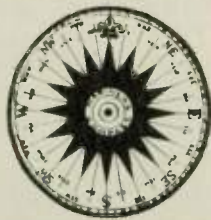
BASIC LIST

AYDIN MODEL NUMBER AMA-	FREQUENCY RANGE (GHz)	GAIN dB MIN.	GAIN VAR. ±dB MAX.	NOISE FIGURE dB MAX.	VSWR MAX. IN/OUT	SATURATED POWER OUT dBm	SATURATED POWER VAR. dB	BIAS	
								VOLTS	mA
1020L	1.0-2.0	40	1.0	3.5	2.0	+12	1.0	+12	220
1530L	1.5-3.0	40	1.0	4.0	2.0	+12	1.0	+12	220
2040L	2.0-4.0	40	1.0	4.5	2.0	+12	1.0	+12	220
2652L	2.6-5.2	40	1.0	5.5	2.0	+12	1.0	+12	270
2756L	2.7-5.6	65	1.0	6.0	2.0	+12	1.0	+12	400
2756L1	2.7-5.6	40	1.0	6.0	2.0	+12	1.0	+12	300
2060L	2.0-6.0	40	2.0	6.5	2.0	+12	1.0	+12	350
4080L	4.0-8.0	40	1.0	6.5	2.0	+12	1.0	+12	300
70120L	7.0-12.0	40	1.5	8.0	2.0	+12	1.5	+12	400
80120L	8.0-12.0	40	1.5	8.0	2.0	+12	1.5	+12	400
80160L	8.0-16.0	30	2.0	10.0	2.0	+12	1.5	+12	450
80180L	8.0-18.0	30	2.5	10.0	2.0	+12	2.0	+12	500
60180L	6.0-18.0	30	2.5	10.0	2.0	+12	±2.0	+12	550

1. Connectors are SMA-F, unless otherwise requested.
2. Base plate temperature is 0° to 50° C.
3. Many other performance characteristics plus a wide range of options are available.


AYDIN MICROWAVE
 IN TOUCH WITH THE WORLD

Around the Circuit



PERSONNEL

Recent management appointments include **Ronald Richards** to the post of Marketing Manager at Instruments for Industry, Inc. and **Barry Hampton** as Applications Mgr., Microwave Operations at Fujitsu Microelectronics. . . Filling the new position of V.P. of Advanced Systems and Technology at the Applied Technology Div. of Itek Corp. is **Dr. Eckhard J. Schulz**. . . Microdyne Corp. has promoted **George A. Bell** from Dir. of Marketing to V.P., Marketing, Satellite TV Products; **Richard B. Elsea** from Dir. of Marketing to V.P., Marketing, Telemetry Products and **William E. Drouillard**, from Gen. Mgr. to V.P. of Antennas for Communications. . . Narda Microwave Corp. appointed **Henry Suri** as Customer Service Administrator. . . **James H. Hamm** was appointed as Materials Research Corp.'s V.P. for the Ceramic Substrate and Process Equipment Divisions. . . **Arnold Newton** has joined International Microwave Corp. in the new position of Dir. of Engineering and **Jack C. Paterno** was elected V.P. of Finance. . . Texscan Corp. announced the promotion of **Raleigh B. Stelle III** from Sales Manager for the Indianapolis office to Vice President and **Norrie Westcott** from Operations Manager for the Indianapolis manufacturing facility to Vice President. . . General Microwave Corp. appointed **Leonard Trugman** as V.P.-Manufacturing. . . Microdyne Corp. appointed **Kenneth B. Boothe** as International Sales Manager. . . **Louis J. Nielsen** was appointed Components Marketing Mgr. at Anaren Microwave, Inc.

CONTRACTS

Warner Robins Air Logistics Center awarded a \$2M contract to **Narda Microwave Corp.** for GaAs FET amplifiers to be used to upgrade the ALR-20A EW system on the B-52. . . **E-Systems, Inc.** was granted a \$13.1M award from US Navy's Sea Systems Command for the production of 7 AN/SYR-1 systems and spares for downlink communications tracking sets used with the extended range Standard Missile-Two. . . **American Electronic Labs** announced the receipt of a \$6.4M USAF award for production of RF Amplifiers and antennas for the EF-111A aircraft. . . AT&T Long Lines Department granted **California Microwave, Inc.** a \$797K order for 2 satellite communications earth stations to be provided by CM's subsidiary, **Satellite Transmission Systems, Inc.** . . **Wold Communications, Inc.**, a subsidiary of **Robert Wold Co. Inc.**, granted a contract valued at over \$4M to **Microdyne Corp.** to manufacture and install at least 100 satellite receive only earth stations, to add to the 50-plus 7-meter stations ordered last year. . . **RCA Missile and Surface Radar** received a \$109.7M contract from the US Navy for purchase of long lead items for the third and fourth AEGIS Weapon Systems. . . **Hughes Aircraft Co. Ground Systems Group**, awarded a subcontract valued at \$500K to **TRAK Microwave Corp.**, the **Tech-Sym Corp.** subsidiary, for transistor power amplifiers for JTIDS. . . AT&T

Co. has entered into a national supply contract with **Avantek, Inc.** for its DR2C and DR2R 2 GHz digital microwave radio equipment and associated services.

INDUSTRY NEWS

KW Engineering has occupied a new 8300 sq. ft. facility at 4565 Ruffner Street in San Diego, CA. . . **Tecom Industries** announced the acquisition of the **Technetics Div. of Duntech Industries, Inc.**, Las Cruces, NM for an undisclosed sum of cash. The Technetics Div. manufactures electrically small, very wideband and compact antennas particularly suited to mobile applications. . . **K & L Microwave** has occupied an addition to their facility which brings total manufacturing space to 38,000 sq. ft. . . **Alpha Industries, Inc.** has agreed in principle to acquire **Trans-Tech, Inc.** of Gaithersburg, MD, a producer of mw ceramic and ferrite materials, and **Central Microwave Co.**, of St. Charles, MO., a manufacturer of GaAs mm and mw signal sources. Alpha plans to issue approximately 350K new shares of common stock in exchange for the outstanding stock of the two companies. . . Board of Directors of **Aeroflex** (Plainview, NY) and **Comtech Telecommunications Corp.** (Smithtown, NY) jointly announced the approval for negotiation of a merger of the two companies on the basis of 1 common share of Aeroflex for each 3 common shares of Comtech. . . **Diamond Antenna & Microwave Corp.**, incorporated by its current President, **Albert S. Hovannesian**, in 1956, is celebrating its 25th year as a designer and manufacturer of rotary joints, antennas and passive components. . . All outstanding stock of **Cincinnati Electronics Corp.**, was purchased by **Marconi Electronics Inc.**, a subsidiary of the **General Electric Co. Ltd.**

NEW MARKET ENTRY

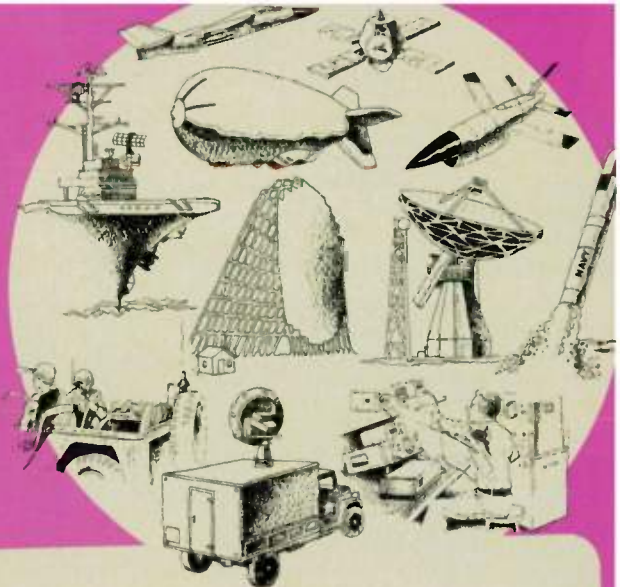
PMH, Inc. has been founded to supply precision machines housings for EMI/RFI suppression applications. The new company will rely heavily on computer-aided machining to produce close tolerance, custom designs. Contact: **Robert Price**, 410 Coles Circle, Salisbury, MD. Tel: (301) 749-2424.

FINANCIAL NEWS

Comtech Telecommunications Corp. reported second quarter results for the period ended January 31, 1981 of net sales of \$10.9M, net income of \$2.0M or 55¢ per share. This compares with 1980 quarterly net sales of \$5.6M, net loss of \$377K or 11¢ per share. . . **Narda Microwave Corporation** reported net sales for the third quarter ended March 31, 1981 of \$6.5M, net income of \$523K and earnings per share of 34¢. In the 1980 third quarter, sales totaled \$5.7M, net income was \$304K, 27¢ per share. . . **Raytheon Co.** reported a net income of \$73.2M for the first quarter ended March 29, 1981, sales of \$1.3B and earnings per share of \$1.75. In the comparable 1980 quarter, net income was \$65.2M, \$1.57 per share on net sales of \$1.197B. . . For the year ended November 2, 1980, **Microdyne Corporation** reported sales of \$23.7M, net income of \$3.2M or 77¢ per share. . . This compares with 1979 year-end net sales of \$17.5M, net income of \$2.9M or 79¢ per share. . . **EPSCO, Inc.**'s results from operations for the year ended December 31, 1980 were net sales of \$16M, net income of \$935K or 96¢ per share. This compares with 1979 year-end net sales of \$13.3M, net income from operations of \$876K, 94¢ per share. ☞

Looking for the world leader in solid state RF/Microwave power?

You've found it!



RF/Microwave products for telecommunications, defense electronics, laboratory instrumentation and test—frequencies from less than 1 MHz up to 8.4 GHz—power outputs from less than 1 watt up to several kilowatts—all solid state—that's the world of MPD. It's a world that's getting wider every day, with a constantly growing spectrum of commercial, industrial and military applications:

Space Satellite Amplifiers

- Class A linear or Class C
- 1500 to 2300 MHz
- "Space-Qualified" transmitters

Satellite Ground Stations

- GaAs FET power amplifiers, up to 8.4 GHz
- 1 and 5 MHz distribution amplifiers, up to 26 outputs
- FM carrier baseband video distribution amplifiers
- IF amplifiers, 70, 700 and 1100 MHz

Terrestrial Microwave Amplifiers

- Microwave LOS, 100 watts
- High power troposcatter, L-band, 1000 watts
- Microwave and UHF radio relay, 1000 watts

Broadcast

- UHF/VHF color TV transmitters, up to 1.5 KW peak synch
- Airborne TV visual/sound power amplifiers
- FCC type-accepted driver amplifiers
- UHF TV internal 3-tone amplifiers

Avionics

- FAA and MIL TACAN transmitter systems— power amplifiers, modulators, synthesizers, power supplies
- L-band digital transmitters (JTIDS)
- Data link transmitters
- Up/down converters
- Airborne pulse amplifiers

Missile Systems

- Command/destroy transmitters
- Guided weapon data link amplifiers
- Military drone transmitters

Radar Amplifiers

- L-band transmitters
- S-band pulse drivers for 3-D radar
- Shipboard drivers for AN/SPS-48 radars

Electronic Warfare

- Communication jammers
- Class A linear power amplifiers
- Linear AB wideband jammers
- Jamming simulators

Military Communications Amplifiers

- Long-pulse data links
- Communication command links
- UHF transceiver amplifiers/modulators
- MIL RF power boosters (ECCM)

Laboratory Instrumentation/Test

- Class A linear amplifiers
- RFI/EMI test amplifiers
- Power meter calibration systems
- Commercial and MIL power supplies— high efficiency, compact packaging, up to 6000 watts
- DC-DC converters



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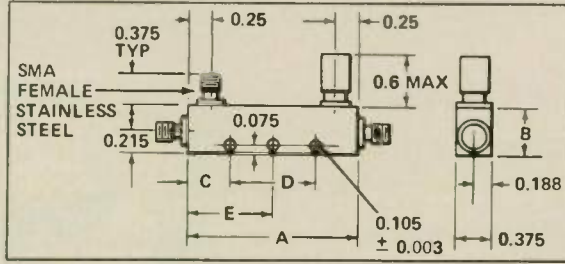
World Radio History
CIRCLE 85 ON READER SERVICE CARD

THE MIDISCO CONNECTION

DIRECTIONAL COUPLERS 0.5 to 18 GHz

Many of the couplers shown exceed a frequency bandwidth and exhibit excellent flatness and directivity characteristics in small packages. Multi-octave units are available.

MIDISCO offers a line of standard, multi-octave, dual directional and high power units. These couplers exhibit excellent flatness and directivity characteristics in small packages. They are available from stock.



OUTLINE DRAWING NUMBER	A	B	C	D	E
1	3.10	0.50	0.80	1.50	—
2	3.10	0.55	0.80	1.50	—
3	1.80	0.50	0.43	0.94	—
4	1.80	0.55	0.43	0.94	—
5	1.20	0.50	0.43	0.34	—
6	1.20	0.55	0.43	0.34	—
7	1.00	0.50	—	—	0.50
8	1.00	0.55	—	—	0.50
9	1.00	0.60	—	—	0.50



MODEL NO.	FREQUENCY RANGE (GHz)	COUPLING* (dB)	FREQ. SENS. (dB)	INSERTION LOSS (dB)		DIREC. TIVITY (dB MIN.)	VSWR MAX.		POWER			OUTLINE DWG. NO.
				EXCL. CPLD PWR.	TRUE		PRIMARY LINE	SECONDARY LINE	AVERAGE INCIDENT (W)	AVG. REFLD. (W)	PEAK (kW)	
MDC6223-6	0.5-1.0	6 ± 1.00	± 0.60	0.20	1.80	23	1.15	1.15	50	2	3	1
MDC6223-10	0.5-1.0	10 ± 1.25	± 0.75	0.20	0.80	25	1.15	1.15	50	5	3	1
MDC6223-20	0.5-1.0	20 ± 1.25	± 0.75	0.20	0.20	25	1.15	1.15	50	50	3	1
MDC6223-30	0.5-1.0	30 ± 1.25	± 0.75	0.20	0.20	25	1.15	1.15	50	50	3	2
MDC6224-6	1.0-2.0	6 ± 1.00	± 0.60	0.20	1.80	25	1.15	1.15	50	2	3	3
MDC6224-10	1.0-2.0	10 ± 1.25	± 0.75	0.20	0.80	25	1.10	1.10	50	5	3	3
MDC6224-20	1.0-2.0	20 ± 1.25	± 0.75	0.20	0.20	27	1.10	1.10	50	50	3	3
MDC6224-30	1.0-2.0	30 ± 1.25	± 0.75	0.20	0.20	27	1.10	1.10	50	50	3	4
MDC6225-6	2.0-4.0	6 ± 1.00	± 0.60	0.20	1.80	22	1.15	1.15	50	2	3	5
MDC6225-10	2.0-4.0	10 ± 1.25	± 0.75	0.20	0.80	22	1.15	1.15	50	5	3	5
MDC6225-20	2.0-4.0	20 ± 1.25	± 0.75	0.20	0.20	22	1.15	1.15	50	50	3	5
MDC6225-30	2.0-4.0	30 ± 1.25	± 0.75	0.20	0.20	22	1.15	1.15	50	50	3	6
MDC6266-6	2.6-5.2	6 ± 1.00	± 0.60	0.25	1.80	18	1.25	1.25	50	2	3	7
MDC6266-10	2.6-5.2	10 ± 1.25	± 0.75	0.25	0.80	20	1.25	1.25	50	5	3	7
MDC6266-20	2.6-5.2	20 ± 1.25	± 0.75	0.25	0.25	20	1.25	1.25	50	50	3	7
MDC6266-30	2.6-5.2	30 ± 1.25	± 0.75	0.25	0.20	20	1.25	1.25	50	50	3	8
MDC6226-6	4.0-8.0	6 ± 1.00	± 0.60	0.25	2.00	18	1.25	1.25	50	2	3	7
MDC6226-10	4.0-8.0	10 ± 1.25	± 0.75	0.25	1.00	20	1.25	1.25	50	5	3	7
MDC6226-20	4.0-8.0	20 ± 1.25	± 0.75	0.25	0.30	20	1.25	1.25	50	50	3	7
MDC6226-30	4.0-8.0	30 ± 1.25	± 0.75	0.25	0.25	20	1.25	1.25	50	50	3	8
MDC6227-6	7.0-12.4	6 ± 1.00	± 0.50	0.40	2.00	15	1.30	1.30	50	2	3	7
MDC6227-10	7.0-12.4	10 ± 1.25	± 0.50	0.40	1.00	17	1.30	1.30	50	5	3	7
MDC6227-20	7.0-12.4	20 ± 1.00	± 0.50	0.30	0.35	17	1.25	1.25	50	50	3	7
MDC6227-30	7.0-12.4	30 ± 1.00	± 0.50	0.30	0.30	17	1.25	1.25	50	50	3	8
MDC6288-6	7.5-16.0	6 ± 1.10	± 0.60	0.60	2.00	12	1.35	1.40	50	2	2	7
MDC6288-10	7.5-16.0	10 ± 1.50	± 0.75	0.60	1.00	12	1.35	1.40	50	5	2	7
MDC6288-20	7.5-16.0	20 ± 1.25	± 0.75	0.50	0.50	15	1.35	1.40	50	50	2	7
MDC6288-30	7.5-16.0	30 ± 1.25	± 0.75	0.50	0.50	15	1.35	1.40	50	50	2	8
MDC6228-6	12.4-18.0	6 ± 1.00	± 0.50	0.30	2.00	15	1.35	1.40	50	2	1	7
MDC6228-10	12.4-18.0	10 ± 1.00	± 0.50	0.30	1.10	15	1.30	1.40	50	5	1	7
MDC6228-20	12.4-18.0	20 ± 1.00	± 0.50	0.30	0.50	15	1.30	1.40	50	50	1	9
MDC6228-30	12.4-18.0	30 ± 1.00	± 0.50	0.30	0.50	15	1.30	1.40	50	50	1	9

*Includes Frequency Sensitivity.

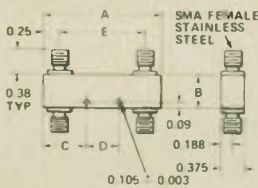


FIGURE 1

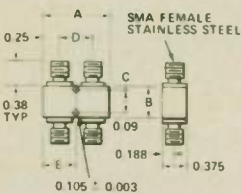


FIGURE 2

HYBRIDS/90° mini

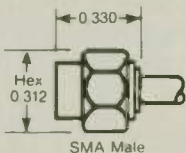
MIDISCO offers this series of 3 dB Miniature Hybrids with excellent electrical performance. These devices are useful as signal and power splitters in both laboratory and system applications. They are available from stock.

MODEL	FREQUENCY RANGE (GHz)	COUPLED (dB) NOM.	DEV. FROM NOM. MAX. (dB)	VSWR MAX.	ISOL. (MIN.) (dB)	INPUT PWR. AVG. (W)	PEAK (kW)	FIG. NO.	A	B	C	D	E
MDC7223	0.5-0.1	3 ^{+0.2} ₋₀	± 0.5	1.25	20	50	5	1	3.06	0.50	0.64	1.780	2.56
MDC7224	1.0-2.0	3 ^{+0.2} ₋₀	± 0.5	1.10	30	50	5	1	1.78	0.50	0.64	1.500	1.28
MDC7225	2.0-4.0	3 ^{+0.2} ₋₀	± 0.5	1.20	22	50	5	1	1.15	0.50	0.314	0.66	0.58
MDC7266	2.6-5.2	3 ^{+0.3} ₋₀	± 0.5	1.25	20	50	5	2	1.00	0.50	0.314	0.50	0.50
MDC7226	4.0-8.0	3 ^{+0.3} ₋₀	± 0.5	1.25	20	50	5	2	1.00	0.50	0.314	0.50	0.50
MDC7227	7.0-12.4	3 ^{+0.4} ₋₀	± 0.5	1.30	18	50	5	2	1.00	0.50	0.314	0.50	0.50
MDC7288	7.5-16.0	3 ^{+0.6} ₋₀	± 0.6	1.35	15	50	5	2	1.00	0.58	0.392	0.50	0.50
MDC7228	12.4-18.0	3 ^{+0.6} ₋₀	± 0.7	1.40	15	50	5	2	1.00	0.58	0.392	0.50	0.50
MDC7267	2.0-8.0	3 ^{+0.8} ₋₀	± 0.6	1.30	17	50	5	1	2.54	0.75	0.64	1.260	2.04

need to hunt another further, the mystery is solved. Midisco has your microwave components: attenuator, cable assembly, coaxial detector & short, directional coupler, hybrid, mixer, phase adjuster, RFI shielded case, inter & intra series adapter, termination, test cable.

SMA SEMI-RIGID MIDGETS

Smallest available, solder type in BeCu or Stainless Steel with passivated or gold-plated coupling nut per MIL-C-39012 0.312D x 0.330. For 141 (RG402 u) cable, order MDC1141 (25 to a bag) \$2 35 ea. For 085 (RG405 u) cable, order MDC1085 (25 to a bag) \$2 95 ea

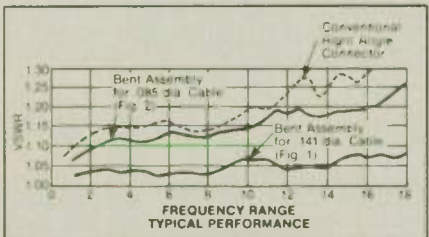
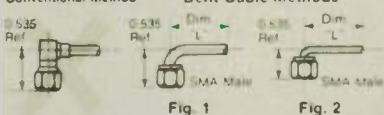


RIGHT ANGLE BENT ASSEMBLY

Get the performance of a straight connector and the dimensions of a right angle plug in one assembly. Order MDC1052 for 0.141 cable or MDC1053 for 0.085 cable. 2" to 8" assemblies, MDC1052 or MDC1053 (10 piece price) \$24.00 each, any length available.

Conventional Method

Bent Cable Methods



Make the proper connection.

call Midisco for a solution (516) 543-4774 or TWX 510-226-7839 and end your search

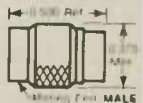


PRECISION SMA TERMINATIONS all stainless steel with 50 ohm impedance and -50° C to +125° C temperature range

Part No.	Av. PWR (Watts)	DC-4GHz	VSWR (max.) 4-8GHz	VSWR (max.) 8-12GHz	12-18GHz	Dimensions L. x W.	Price (Each in quantity of 10)
MDC1075	1/2	1.05	1.10	1.15	1.20	55 x 31D	\$18.00
MDC1075H	2	1.05	1.10	1.15	1.25	69 x 31D	\$28.00
MDC1091M	5	1.05	1.10	1.15	1.20	32 x 31D	\$18.00
MDC1091MX	1/2	1.03	1.04	1.08	1.12	32 x 31D	\$35.00
MDC1076	1/2	1.05	1.10	1.15	1.20	52 x 31D	\$18.00
MDC1090F	5	1.05	1.30	1.50	—	755 x 5 sq	\$35.00
MDC1090M	5	1.05	1.30	1.50	—	895 x 5 sq	\$35.00

SMA KWIK-SLIP TERMINATIONS

Snaps onto any SMA female connector, rated to 18GHz MDC1075Q stainless steel-gold plated 1 watt average power, 50 ohm imp. 1.05 + 0.010F (GHz) VSWR -55° C to +125° C temp range, 500 cycle (min.) durability, 10 lb axial engagement force, 10 piece price, each \$38, Model MDC1075QC, with bead chain, 10 piece price, each \$48

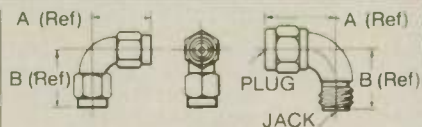


Precision Adapters



Every combination, inter & intra series adapter between N, BNC, TNC, SMA, SMB, SMC and 7mm in straight, bulkhead, panel and bent versions.

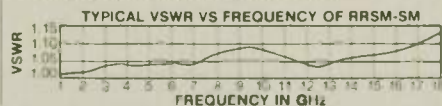
MIDISCO'S Number System makes it so easy: Type N = N, TNC = T, BNC = B, 7mm = 7, SMA = S, Male = M, Female = F. So a TNC male to SMA female is TM-SF. Prefix with B, bulkhead, P, panel, R, right angle, RR, radius right angle.



MIDISCO Radius Adapters consist of a bent section with a one piece center conductor eliminating vibration & shock failures that are normal with solder joints. VSWR to 18 GHz is superior to mitered right angle adapters.

Type	PART NO.	A	B
N Male - N Male	RRNM-NM	1.064	1.064
N Male - N Female	RRNM-NF	1.109	1.000
SMA Male - SMA Male	RRSM-SM	0.550	0.550
SMA Male - SMA Female	RRSM-SF	0.550	0.460
TNC Male - TNC Male	RRTM-TM	1.060	1.060
TNC Male - TNC Female	RRTM-TF	0.983	1.060

\$45 each and 1 \$25 each in 1 to 9 quantity



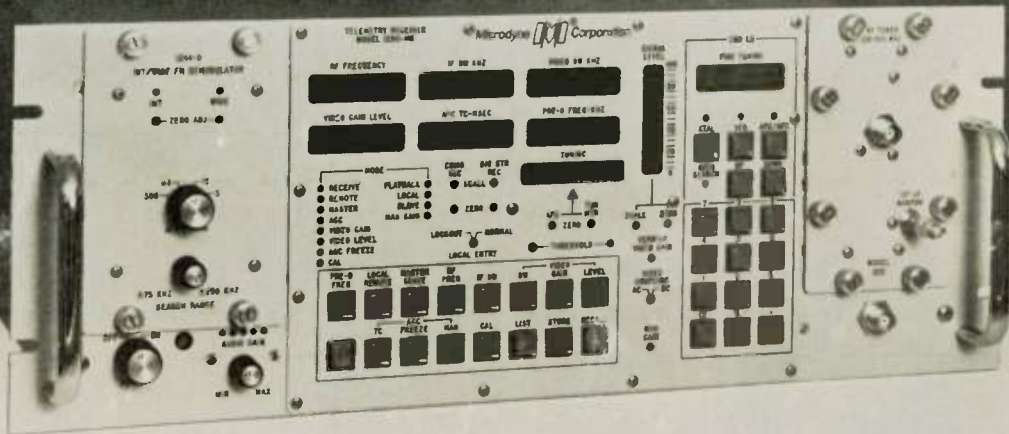
Make the proper connection,
call Midisco for a solution
(516) 543-4774 or
TWX 510-226-7839
and end your search.

MIDISCO

See all the MIDISCO COMPONENTS at Booth #526, MTT-S.

61 MALL DRIVE,
COMMACK,
NEW YORK 1172

the telemetry receiver your computer really understands



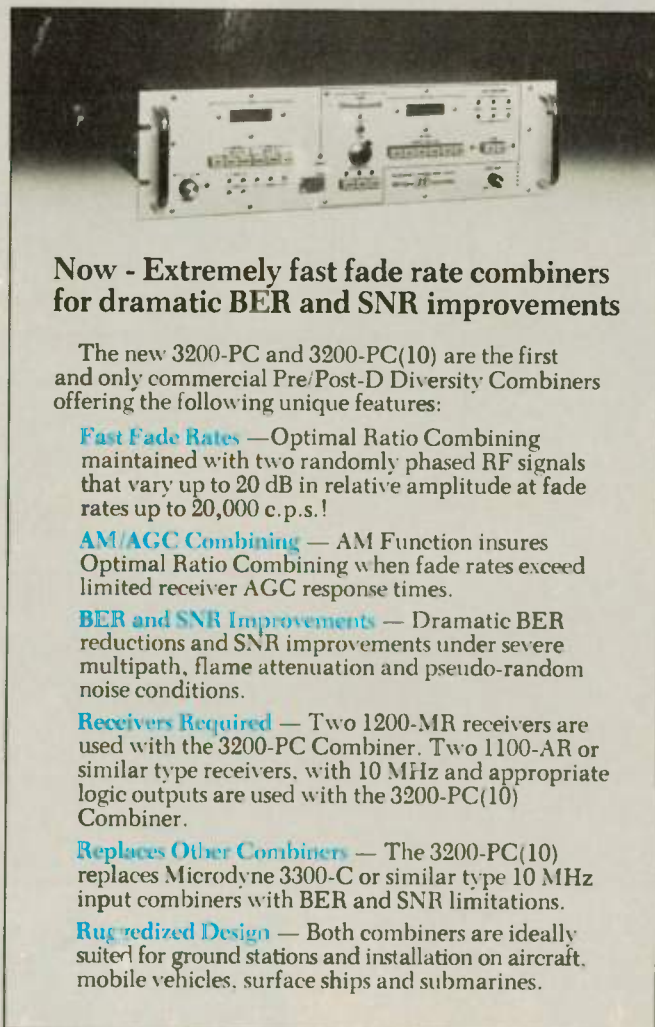
Microdyne's New 1200-MR

Both remotely and manually controllable, the 1200-MR double conversion receiver is the first to anticipate and meet the many telemetry requirements of the 1980's. Consider just some of the features:

- IEEE-488 or optional RS-232 interface
- Keyboard entry for manual control
- Discrete tuning ranges, voltage controlled from 215 MHz to 4.2 GHz
- AM, FM, PM and Bi-Phase Demodulation Modes
- DC to 5 MHz video Baseband response
- Synthesized LO's, phase locked to a common 5 MHz reference, with 1×10^{-7} stability
- Up to seven internal 2nd IF filters, switch selectable with BW's up to 10 MHz
- Ruggedized design allows installation on aircraft, land vehicles, ships and submarines

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MW/MM-Wave Tube and Transmitter Technology Development Contracts at ERADCOM

NELSON J. WILSON

US Army Electronics Technology and Devices Laboratory
Fort Monmouth, NJ

This report will review and update the status and progress of the ERADCOM contractual programs for development of tubes and modulator subsystems.

Varian-Canada under contract DAAB07-78-C-2948 has continued to make excellent progress with development of the four-cavity, extended interaction amplifier (EIA) operating from 94.0-96.0 GHz. The initial electromagnet (EM) focused amplifier has been packaged with a samarium-cobalt magnet and weighs approximately 6.6 kg. Operation at 2.4 kW, 43-46 dB gain, 19% efficiency duplicates the earlier EM performance. The tube is mechanically tunable from 94-96 GHz with an electronic bandwidth of 180 MHz at the 1 dB power point. The maximum average power obtained to date is 50 watts under liquid-cooled operation. Typical average power is approximately 30 watts using a 0.4 μ sec pulse width. Excellent stability was demonstrated in a recent pulsed evaluation at Electronics Technology and Devices Laboratory (ET&DL). Plans are to reduce the overall weight, provide for mod-anode or grid controlled pulsing and to extend the design to 220 GHz.

Hughes Electron Dynamics Division (EDD) under contract DAAB07-78-C-3015 has completed the design for a 100 watt peak power periodic permanent magnet (PPM) focused coupled-cavity TWT (*Microwave Journal*, July 1980 p. 40). An electron beam tester was fabricated and evaluated. The desired 95% (minimum) beam transmission was not achieved and a refinement of the magnet stack was completed. Plans are to fabricate a new electron gun and assemble the available circuit parts with a new magnet stack to obtain the first 94 GHz model.

Hughes EDD has completed contract DAAB07-77-C-2713, "Linear 1 kW Multitone TROSCATTER TWT,"

by constructing a high power TWT utilizing a four-stage depressed collector to enhance efficiency. This PPM focused coupled-cavity TWT operates in a non-saturated drive region to limit intermodulation distortion to -20 dB. The massive copper four-stage depressed collector contains machined relief slots to minimize thermal expansion while collecting a total power of 4.3 kW. This 1.0 kW amplifier operates across the 4.4-5.0 GHz frequency band at a minimum gain of 40 dB. Final evaluation will be completed by Sept. 1981.

Parallel advanced development contracts at Varian (DAAB07-78-C-3007) and Hughes (DAAB07-78-C-3008) have resulted in obtaining improved 125 kW peak power transmitter tubes for ground radars operating in S Band. Improvements in amplitude linearity, reduced cathode temperature and increased resistance to load mismatch (1.5:1 min.) have been demonstrated during evaluation on contract DAAB07-78-C-3010, Transmitter Test Bed Program. A microprocessor-controlled transmitter has operated in excess of 400 hours with no degradation in either tube or transmitter performance. Evaluation will continue until 1,000 hours of operation has been attained on each of two tubes.

In companion programs at Litton (DAAB07-80-C-0294) and Hughes (DAAB07-80-C-0295), development is underway to improve an X-Band, 25 kW peak power traveling-wave tube. Tungsten-iridium cathodes will be employed to reduce cathode temperature and provide for longer life. Room temperature (passive) getters will be introduced to inhibit the effects of gases evolved during extended shelf life. An external network will be incorporated with each tube to eliminate improper tube voltage settings, and heater and beam current timers are also being

provided to assist in gathering and evaluating life data in the field. The development includes demonstration of a 1,500-hour life test when operated in a microprocessor controlled transmitter test bed.

PULSERS AND SWITCH TUBES

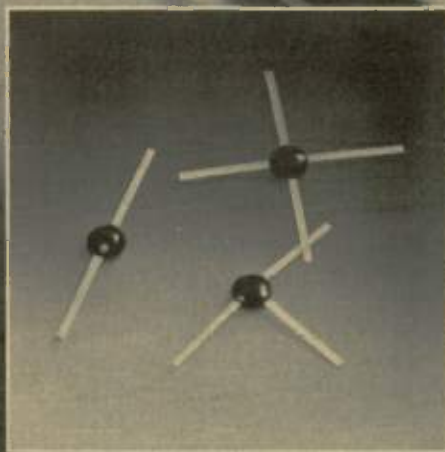
Applications for high resolution radars operating at millimeter wave lengths have led to the development of nanosecond pulse lengths at the 2-3 kV level. Under contract DAAB07-78-C-2991, Hughes Ground System Division has extended the capability of their 1.0 kV avalanche transistor pulser. Pulse step-up of 2:1 and 3:1 has been achieved by fabricating coaxial line transformers wrapped on microwave ferrite core. Using a 50 ohm input impedance and terminating the transformer into a 200 Ω load resulted in obtaining a uniform pulse at the 2.0 kV level, while maintaining a 0.4 nanosecond rise time. In order to obtain higher current levels, Hughes will concentrate on designing transformers with lower characteristic impedance levels. The 2-3 kV pulse will provide sufficient pulse drive for gridded extended interaction amplifiers and TWT's.

Instant start thyratrons provide a technology that reduces equipment turn-on time by eliminating the relatively long term warm-up usually associated with thyatron pulsers. The resultant performance and small size of the instant start thyratrons is being exploited for short pulse millimeter wave transmitters. EG&G under contract DAAK20-80-C-0282 is developing a modified HY3013L to provide 25 kV pulses in a low inductance pulser design. The first model thyatron was aged at a pulse voltage of 26 kV with an anode fall time of 2 nanoseconds. Additional characterization testing will be conducted to measure recovery time. ☐

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INTRODUCTION

The MLS (Microwave Landing System) is presently being developed in several countries, including USA and France. The MLS is the international landing system of the future, and will gradually replace the ILS (Instrument Landing System) used at airfields all through the world.

The MLS characteristics are defined by ICAO,* as well for the

- complete position information in 3 coordinates
- wide angular coverage
- good performances even in difficult sites (mountains, etc.)
- easy implantation due to the relative small size of its antennas
- presence of a ground-to-air data link.

Azimuth: it generates the azimuth angle with reference to the runway axis.

Elevation: it generates the elevation angle with reference to the horizontal.

Distance: it generates the distance to a reference point along the runway.

It appears that the MLS gives to the aircraft its position in a quasi-spherical coordinate system related to the runway. The aircraft is therefore able to follow its desired landing path, by determining at each moment in three coordinates its position error compared to the desired path. The main difference between ILS and MLS appears now: ILS gives a position error in relation to a fixed straight landing path, corresponding to $\theta = 0^\circ$ (azimuth) and $\varphi = \varphi_0$ (3° or 2.5° , elevation) angles, while MLS enables an aircraft to follow every desired approach and landing path, including curved approaches.

The distance information is generated by a DME (Distance Measuring Equipment) including an airborne interrogator and a ground transponder. This DME may be either the conventional DME-N as defined presently by ICAO, or a precision DME which characteristics are now discussed at ICAO, but which shall be interoperable with DME-N.

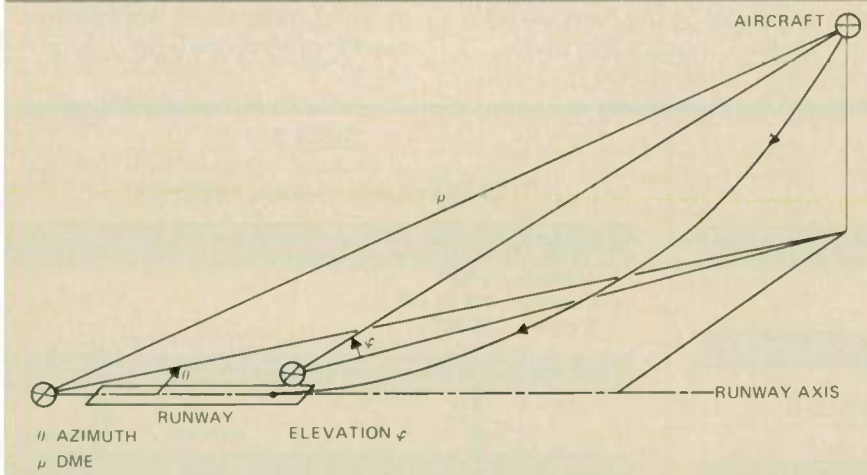


Fig. 1 MLS geometry.

performance and for the technical characteristics of the transmission. The MLS, as an advanced landing system, will offer the same integrity and guidance accuracy as the modern ILS's; moreover, it will offer several other advantages:

MAIN FEATURES OF THE MLS

The Microwave Landing System, when installed along a runway, gives to an aircraft its actual position related to the runway and therefore enables it to land even in poor visibility conditions.

The MLS system consists of the three following functions (see Figure 1):

*ICAO: International Civil Aviation Organization

Scanning Beam Antenna

This is the main part of the equipment as it defines the guidance accuracy. A block diagram of the antenna is given in Figure 6. A phased array, it includes:

- A *power divider*, which splits the transmitter power to the various phase shifters and waveguides. It gives an amplitude tapering on the array for obtaining a low sidelobes level and is realized in microstrip technology.
- *Phase-shifters*, which generate the scanning beam, is explained in Appendix 1. The phase-shifters are controlled at each time so as to generate a beam in the desired direction. By suitable switching of the phase-shifters, it is possible to scan the beam at constant angular speed in a TO-FRO sequence, as required. The phase-shifters are 4 bits digital phase-shifters using PIN diodes, which sample a 360° phase-shift into 16 steps of 22.5°. They are realized in stripline technology. The RF part of the phase-shifters presently used are manufactured by Microwave Associates. The drivers include a BIT (Built in Test) which enables to detect a failure of a single PIN diode, for monitoring and maintenance purposes.
- *Slotted waveguides*, which are the radiating elements. Their coverage defines the coverage of the azimuth antenna. But their main purpose is to avoid reflections of RF waves on the ground in front of the antenna, and to avoid guidance noise. Therefore, their elevation pat-

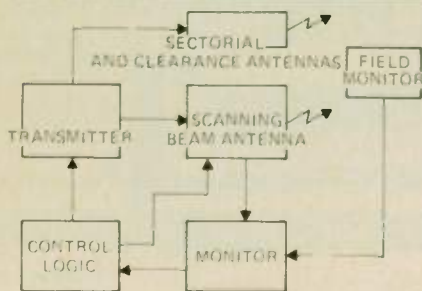


Fig. 5 Azimuth station – block diagram.

TABLE 3
MAIN TECHNICAL CHARACTERISTICS OF MLS 800 A 30

COVERAGE	
– Azimuth	± 40°
– Elevation	0.9° to 15°
– Distance	20 NM
•	Proportional azimuth information over ± 10°
•	Clearance information between 10° and 40° (fly right or fly left information)
ACCURACY AT RUNWAY THRESHOLD	
•	Bias error ± 3 m
•	Noise error (CMN at 2σ) ± 1 m
FORMAT: as required by ICAO standards:	
•	Frequency: 5031 to 5090.7 MHz (200 channels with 300 kHz spacing)
•	Time division multiplex between Azimuth, Elevation and Basic data informations
•	Average rates: Azimuth: 39 Hz Elevation: 39 Hz Basic data (variable according data)
•	Modulation of preamble and basic data: DPSK at 15.625 kHz.
MECHANICAL CHARACTERISTICS	
•	Approximate dimensions of Azimuth station:
– height	2 m
– max height above ground	2.8 m
– width	2 m
– depth	1 m
•	Approximate weight of Azimuth station: 350 kgs.

tern presents a sharp cutoff around horizontal (better than 7 dB/deg) and low sidelobes at negative elevation angles. Slotted waveguides were chosen as radiating elements because they enable good radiation patterns with little hardware, but they require elaborated computer programs for design and precision tooling of the slots. To obtain vertical polari-

zation of the RF waves, the slots are on the small side of the waveguide.

Transmitter

The transmitter includes the following sub-assemblies:

- A frequency synthesizer generating the 200 MLS channels spaced at 300 kHz between 5031 and 5090.7 MHz. It is a phase-locked loop frequency

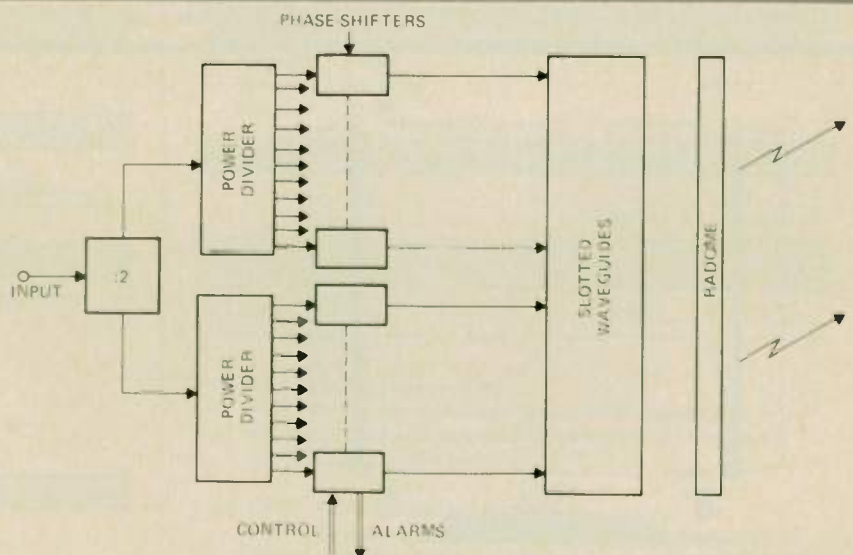


Fig. 6 Scanning beam antenna.

(continued on page 118)

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synthesizer of classical conception.

- A 0-180° phase modulator, generating the DPSK phase modulation of the preamble and of the data. It is a PIN diode phase shifter.
- A power amplifier, with a 20 W output level. A TWT (traveling-wave tube) amplifier is presently used for the prototype. Later on, solid state power amplifiers will be used when 5 GHz power transistors will be available with sufficient output power; at present time, the maximum output power of a single transistor is around 4 W.

Control Logic

The control logic includes two major sub-assemblies:

- **Scanning antenna control logic:** These circuits control at each time the position of each phase-shifter, according to an optimized sequence to ensure soft scanning of the beam and correct beam pointing. They are realized with conventional wired logic, as conventional microprocessors are not fast enough for this purpose.
- **Time sequence control microprocessor:** It generates the time sequence of MLS, with time sharing of all various functions. According to its complexity, it is generated by a microprocessor. The microprocessor drives the various sub-assemblies of the station: phase modulator, antenna switch, control logic, etc.

Sectorial and Clearance Antennas

The Azimuth station includes other antennas, also realized with slotted waveguides.

- **Sectorial antenna:** This antenna radiates the DPSK messages (preamble and data) in the MLS coverage.
- **Clearance antennas:** They radiate "fly left" and "fly right" information outside the proportional coverage (between 10 and 40°).

A PIN diode switch enables the selection of the radiating antenna.

Monitoring

The monitoring system is single in the MLS 800. It includes both internal and external monitoring with a field monitor. The field monitor checks the main parameters radiated by the azimuth station and performs an angular measurement in the field and generates an alarm to stop the radiation if the accuracy limits are exceeded. It also performs power level monitoring for both sectorial and scanning beam transmissions and includes a re-

ceiving horn, a Schottky detector, a video amplifier and measurement circuits.

The internal monitor checks the DPSK modulation and the time sequence, to ensure that the MLS message has the characteristics defined by ICAO. These parameters may be monitored internally without loss of integrity. Instrument includes a phase detector and a microprocessor which monitors at each time the characteristics of the time sequence.

**APPENDIX I
PRINCIPLE OF THE
SCANNING BEAM ANTENNA**

The principle of operation of a scanning beam antenna is described in Figure 7: as said before, a phased array is used, the block diagram of which is given in Figure 6; electronic scanning is obtained through digital phase-shifters.

The power divider splits the power to the various phase-shifters and radiators; to obtain a pattern with low sidelobes, unequal power splitting is implemented in order to realize an amplitude tapering on the array. Several mathematical models may be used for the amplitude tapering: Gaussian, Tchebycheff, Taylor; they enable theoretically sidelobes as low as -30 to -40 dB. The general shape of amplitude tapering is given in Figure 7 (in the power divider).

Phase-shifters generate the electronic scanning of the main lobe of the antenna; according to antennas theory, the angular direction θ_0 of the maximum of the main lobe is perpendicular to an "equiphase" plane, which means a plane in space where waves have the same phase. The phase-shift-

ers enable to generate such an equiphase plane.

Assumptions generally made in the design of a phased array are:

- symmetrical amplitude distribution
- uniform spacing d of radiators
- anti-symmetrical phase distribution: two symmetrical phase-shifters have opposite phase
- the outputs of power divider are all in phase.

Consider two phase-shifters symmetrical in the array, called N on Figure 7 ($N = 1$ for the two central phase-shifters). To generate an equiphase plane, they must generate opposite $\pm \Delta\varphi$ phase-shifts so that electrical lengths $PQ = RS$. A simple geometrical calculation shows that, for a pointing angle θ_0 , theoretical value of $\Delta\varphi$ is:

$$\Delta\varphi = \pi (2N - 1) \frac{d}{\lambda} \sin \theta_0 \quad (1)$$

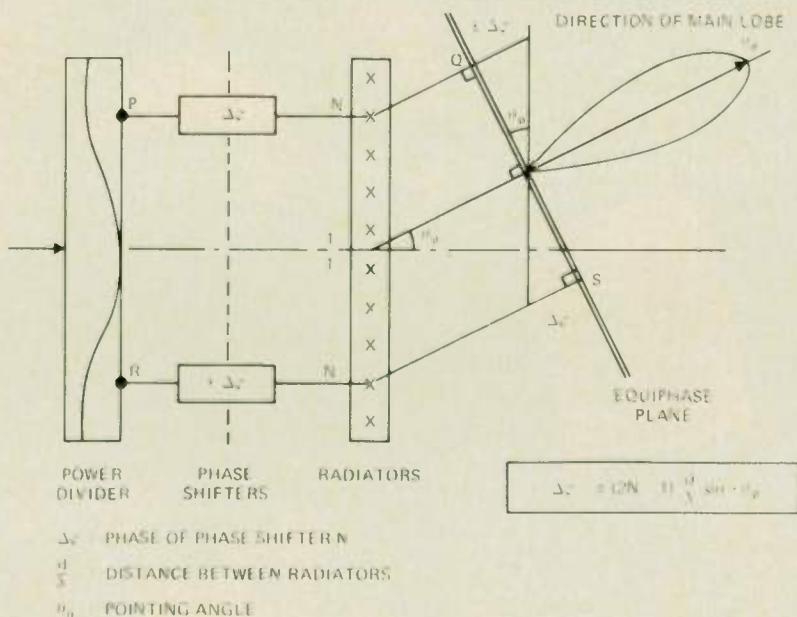


Fig. 7 Principle of the phased array.

(continued on page 120)

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where:

- N = number of considered phase shifter
(with reference to the center)
 d = spacing of radiators
 λ = wavelength
 θ_0 = pointing angle (angle of the maximum of the main lobe)

The equation of the antenna pattern is then (as a function of angle θ):

$$F(\theta) = \sum_{N=1}^{N_0} A(N) \cos [(2N - 1) (\alpha - \alpha_0)] \quad (2)$$

where:

$$\alpha = \pi \frac{d}{\lambda} \sin \theta \quad \theta \text{ angle}$$

$$\alpha_0 = \pi \frac{d}{\lambda} \sin \theta_0 \quad \theta_0 \text{ maximum}$$

$A(N)$ amplitude of radiators N .

The way of generating a scanning beam includes: suitable control of the phase-shifters is made so as to scan angle θ_0 with a constant angular speed. θ_0 is sampled with a small step at successive values $\theta_1, \theta_2, \theta_3$ etc. and for each value θ_i , phase shifters are positioned according Equation 1 to the required value.

When using digital phase-shifters, the exact value of $\Delta\varphi$ given by Equation 1 cannot be realized, and a small phase error occurs on each radiator. The consequence of this phase error is to raise the level of the sidelobes and to give a small pointing error. The control of the phase-shifters must therefore be designed to keep those two effects to a low level.



Bruno Létouart received an electronic, engineer diploma in 1970 from the "École Supérieure d'Electricité" in France. He joined in 1970 Thomson-CSF, Division DRS-TV7, in France. This division deals with radars, ATC and NAVAIDS for both civil and military applications. Mon. Létouart worked on MLS since 1970, and also on various radar projects. He is now at Thomson-CSF in charge of the MLS program, for studying and manufacturing the MLS ground equipments. ☛

quired in the system, one to prevent pulse breakthrough and the other to eliminate excess noise. Both of these gates employed PIN diode switches. A further PIN diode was used in conjunction with a temperature-sensitive resistor for thermal compensation of the delay line attenuation and for overall trimming.

As one or other of the gates was always closed, fast diodes were not required, so that use could be made of large robust devices offering ease of assembly, high switching performance and reliability and the total current demand was constant at substantially 50 mA independent of prf. Attenuation variations were kept to ± 2 dB over 500 MHz bandwidth and a temperature variation of 80°C.

The drive circuit was fairly conventional, using compact printed circuit board, accepting a slightly early trigger pulse through an opto-isolator and generating the required waveforms.

K. Harris of Microwave Associates, Dunstable, described practical examples of diverse uses of the PIN diode in microwave switching circuits, explaining how the design is determined by the application. For high power usage particular attention had to be paid to thermal resistance, tending to result in large diodes of high capacitance and long minority carrier lifetime, good for slow switching at 20 MHz - 1 GHz but far from ideal at frequencies of 10 GHz and upwards. Conversely, small diodes had limited power capability. Whereas few problems in tradeoffs were involved in many low power applications, optimization of diode parameters was important where high power or fast switching capability was required.

The diverse switching applications described included a 1500 CW switching operating from 20 - 100 MHz, using hermetically sealed CERMACHIP PIN diodes, a T.R./A.T.R. fast X-band switch with 200 W CW power capability, using low thermal resistive plated heat sink diodes and a 60 GHz switch suitable for data link applications at powers of up to 1 W, using unencapsulated plated heat

sink diodes. Discussed also, was an experimental bulk window design for use over 26 - 40 GHz instantaneous bandwidth, with a peak power capability in excess of 10 kW.

The final contribution was from R. N. Bates and M. D. Coleman of the Philips Research Laboratories, Redhill, discussing millimeter-wave finline PIN switches for use in the frequency ranges 30 - 40 GHz and 80 - 90 GHz. Two types of switch were described, the one a two-port arrangement for use as an attenuator or modulator and the other a three-port device for use as a Dicke switch in radiometer receivers.

The circuits discussed were all prepared on thin low dielectric constant substrates, the patterns being defined via conventional printed circuit board techniques and hence potentially cheap to produce. They were mounted in the "E"-plane of a rectangular waveguide and incorporated finline, microstrip and coplanar line.

The two-port switch or attenuator comprised a short section of isolated unilateral finline in rectangular waveguide shunted by two or more PIN beam lead diodes, the fin-line being cosine taper matched. By connecting suitable modulators to the loaded fin-line sections, voltage controlled attenuators and switches were realized. For K_a-band (26.5 - 40 GHz) twin-diode units had given attenuation from 0.5 - 20 dB and quad-diode units correspondingly 1 - 30 dB, with "On" current of 2 mA and 1.4 mA, respectively and "Off" voltage 12 V in each case. Corresponding figures for twin-diode units in O-band (80 - 90 GHz) were 1.7 - 20 dB, 1 mA, 15 V.

Three-port switches employed symmetrical "Y" junctions of three isolated unilateral finlines in rectangular waveguide. Using pairs of diodes across each of the output fin-lines, switching over 30 to 40 GHz was obtained with an insertion loss of 1.5 dB and isolation of 20 dB, corresponding figures over 80 to 90 GHz being 2.5 dB, and 20 dB, the "On" current being 8 mA and "Off" voltage 10 V in each case. ☛

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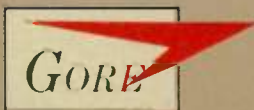
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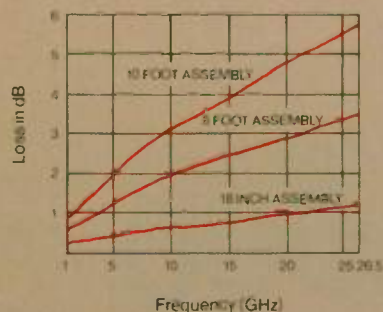
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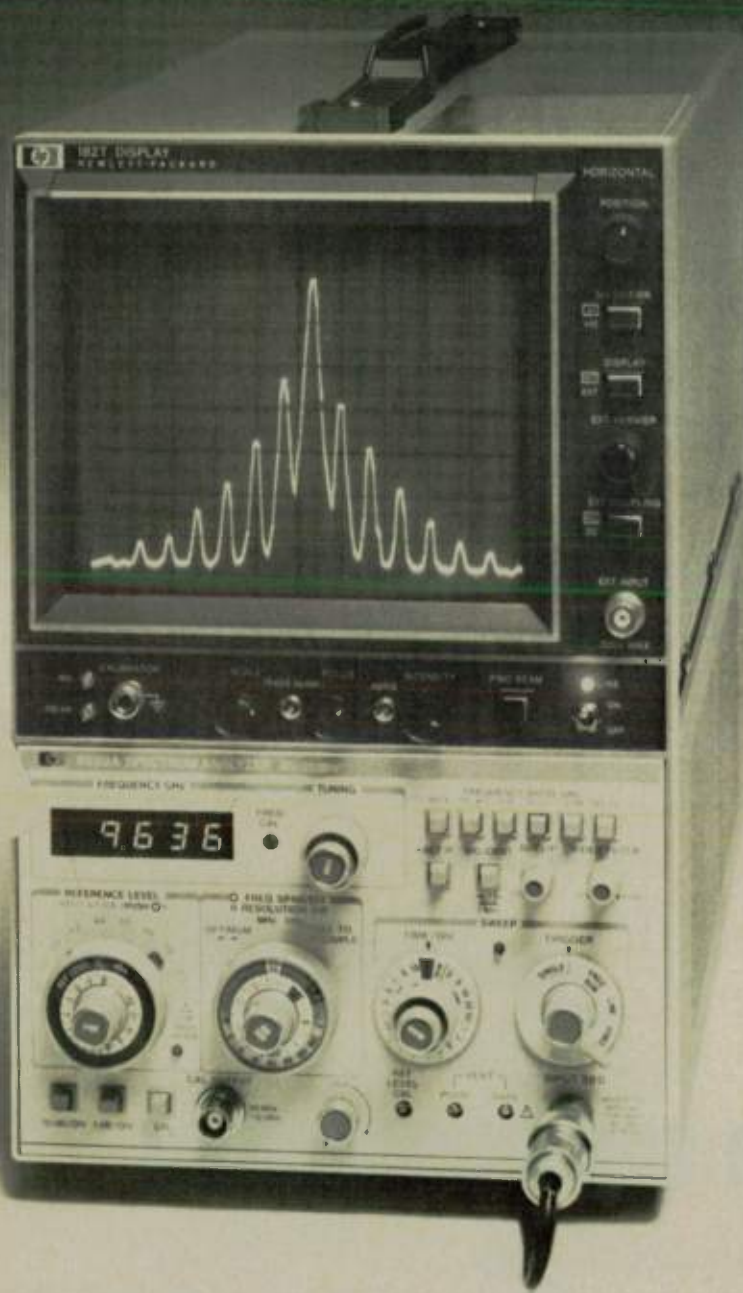
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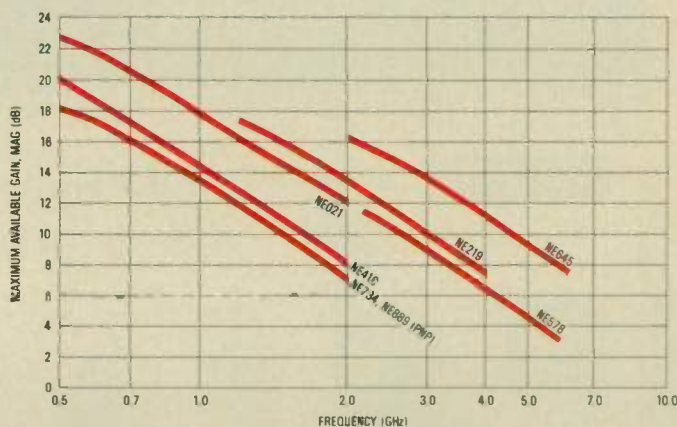
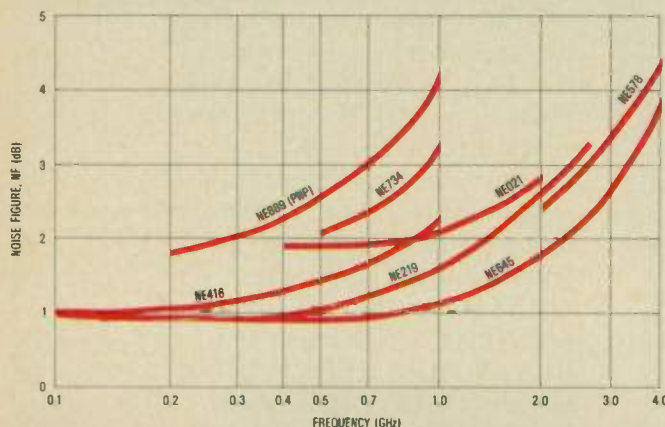
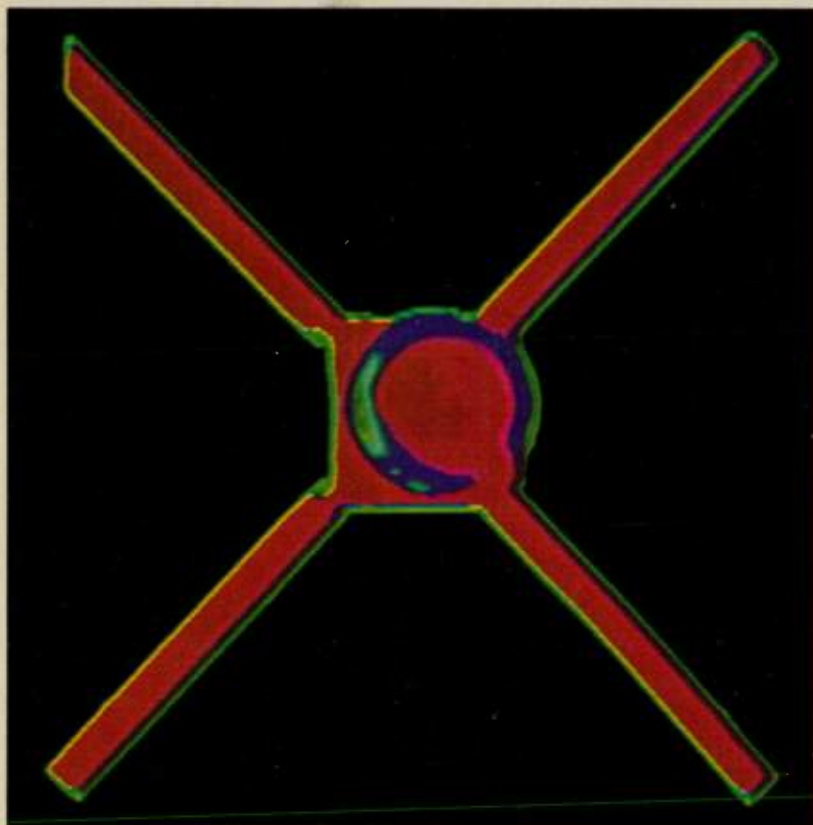
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Graphs of Circuit Elements for the Marchand Balun

J. H. CLOETE

National Institute for Aeronautics and Systems Technology, CSIR, Pretoria, South Africa

INTRODUCTION

The Marchand balun¹ exhibits excellent amplitude balance for bandwidth ratios in excess of a decade. In addition, it can be designed to incorporate transformer action for applications where the load and source resistances differ.² The Marchand balun has been used in a variety of physical implementations, including coaxial and planar, for applications like spiral antennas³ and balanced mixers.⁴

In typical antenna applications the balanced load resistance differs significantly from the source resistance and it is necessary to implement transformer action in the balun in order to achieve an acceptable match. For example, in the case of spiral antennas the balanced load impedance generally has a nominal real part in excess of 100 Ω which must be matched to 50 Ω. A popular design approach to this problem is to design a balun for a 50 Ω load, and then to incorporate a tapered

line section to transform the balanced load resistance to approximately 50 Ω. However, failure to take the interaction between the balun and transformer into account properly can lead to unnecessarily degraded frequency response for broadband designs.

An alternative approach to achieve transformer action is the use of quarter-wave sections instead of a tapered line. This method has the advantage that exact synthesis techniques can be used to include the transformer sections in the balun, thus allowing designs with accurately predictable performance to be implemented. In a recent paper the existence of graphs for the design of a third order balun with a single section quarter-wave output transformer, and a fourth order balun with single section quarter-wave input and output transformers was announced.²

The purpose of this paper is to publish a set of these graphs for the design of the fourth order balun shown in Figure 1. Graphs

for the third order balun will not be published since approximate closed form equations for the element values are available.⁵

Both the third and fourth order baluns have maximum bandwidth ratios in the order of 12:1 due to the high characteristic impedance Z_3 , required for the shunt stub, becoming practically unrealizable. The advantage of the fourth order balun over the third order balun is that for a given load resistance, a significant reduction in passband reflection coefficient can be obtained, depending on the required bandwidth. The improved match is achieved without compromising on the compact physical structure of the Marchand balun, but does involve increased manufacturing expense due to the input quarter-wave line, Z_1 .

THE GRAPHS OF ELEMENT VALUES

The graphs of fourth order Marchand balun circuit element values are given in Figures 2 to 6

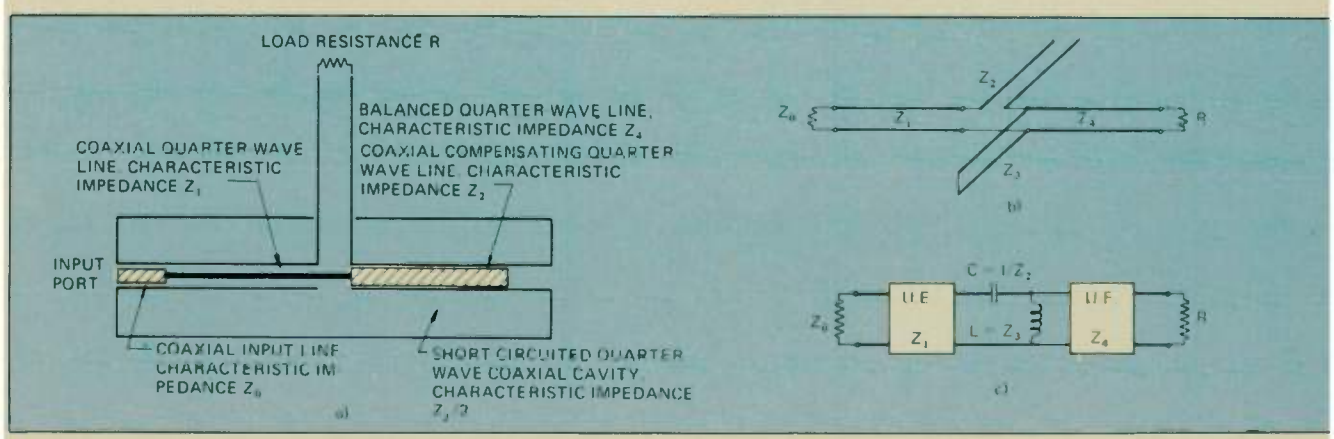


Fig. 1 The fourth order Marchand balun a) Physical circuit; b) Transmission line equivalent circuit; c) Equivalent circuit using distributed element notation.

for five bandwidth ratios ranging from 4:1 to 12:1, and realize the Tchebysheff bandpass response.

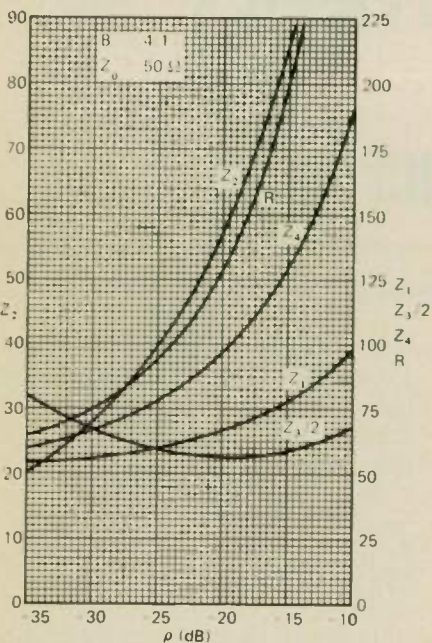


Fig. 2 Graphs of element values for the fourth order balun with source resistance $Z_0 = 50 \Omega$ and bandwidth ratio $B = 4:1$.

For each bandwidth ratio, the element values are plotted as a function of maximum passband reflection coefficient. The notation used in the graphs is consistent with that of Figure 1.

The graphs were produced using the exact, non-redundant method for the synthesis of commensurate length distributed networks described by Horton and Wenzel.⁶ The synthesis program is highly accurate, however, the price to pay for the convenience of the graphs is a loss of accuracy due to read-off errors. In the case of Z_2 , the characteristic impedance of the series open circuit stub, the error should typically be less than $\pm 0.5 \Omega$, and for the other elements the error should typically be less than $\pm 1 \Omega$.

A simple check on the accuracy of the values obtained from the graphs can be made by using the following symmetry equations⁷ for the element values of the balun,

$$Z_1 Z_4 / R = Z_2 Z_3 / R = Z_0$$

where $Z_0 = 50 \Omega$ for the graphs. (see Table II)

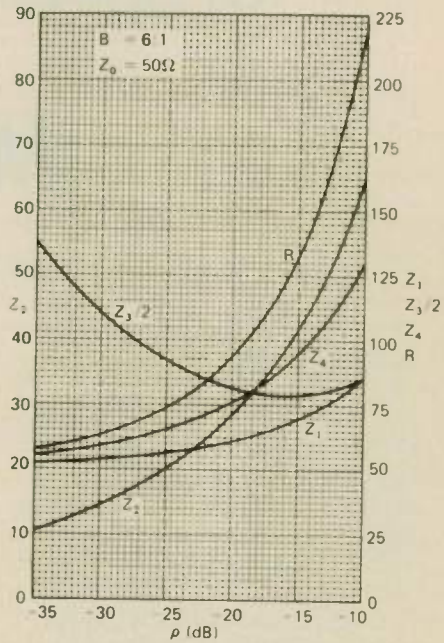


Fig. 3 Graphs of element values for the fourth order balun with source resistance $Z_0 = 50 \Omega$ and bandwidth ratio $B = 6:1$.

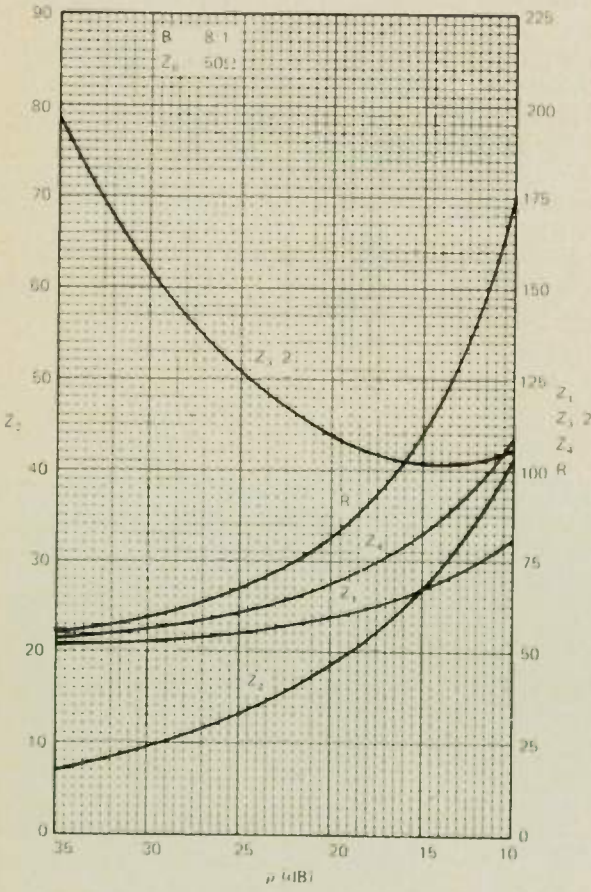


Fig. 4 Graphs of element values for the fourth order balun with source resistance $Z_0 = 50 \Omega$ and bandwidth ratio $B = 8:1$.

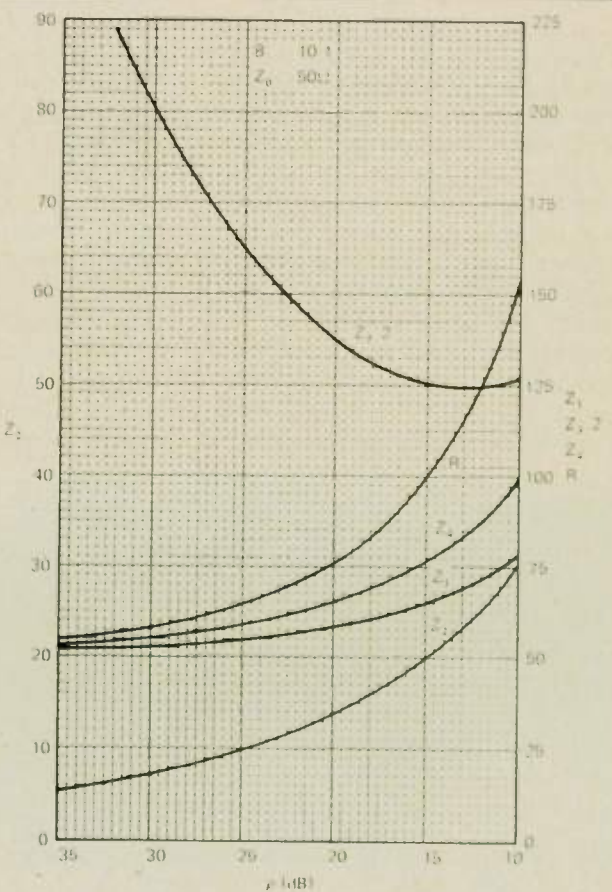


Fig. 5 Graphs of element values for the fourth order balun with source resistance $Z_0 = 50 \Omega$ and bandwidth ratio $B = 10:1$.

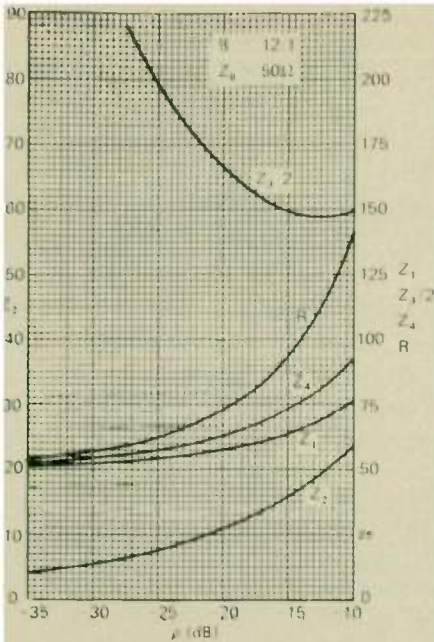


Fig. 6 Graphs of element values for the fourth order balun with source resistance $Z_0 = 50 \Omega$ and bandwidth ratio $B = 12:1$.

DESIGN EXAMPLE

To illustrate the use of the graphs consider the design of a balun with bandwidth ratio 10:1, source resistance 50Ω , and load resistance 100Ω . The element values and maximum passband reflection coefficient are found from Figure 5 and given in Table 1. For comparison, the exact values as computed by the synthesis program are also tabulated. It is seen that the errors are insignificant from a practical viewpoint when manufacturing tolerances are considered. In addition, the

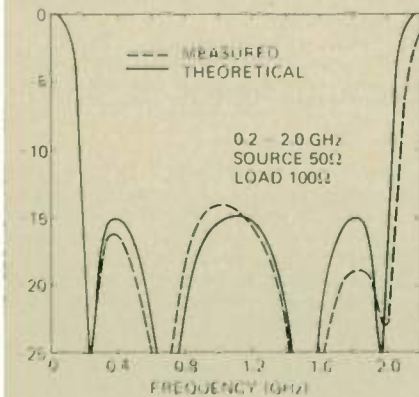


Fig. 7 Measured and theoretical frequency response of an experimental 10:1 bandwidth fourth order Marchand balun with 50Ω source resistance and 100Ω load resistance.

TABLE 1

ELEMENT VALUES FOR A FOURTH ORDER BALUN

	$Z_1 (\Omega)$	$Z_2 (\Omega)$	$Z_3 (\Omega)$	$Z_4 (\Omega)$	ρ (dB)
Exact values	65.14	19.99	250.18	76.75	-14.87
Values from Fig. 5	65.0	20.0	250.0	76.0	-14.9
Error magnitude	0.14	0.01	0.18	0.75	0.03

Bandwidth ratio $B = 10:1$, Source resistance $Z_0 = 50 \Omega$, Load resistance $R = 100 \Omega$

TABLE II

ELEMENT VALUES FOR FOURTH ORDER MARCHAND BALUN AS COMPUTED BY THE SYNTHESIS PROGRAM

FOURTH ORDER MARCHAND BALUN WITH IN AND OUTPUT TRANSFORMERS

BANDWIDTH RATIO = 10

RHO(DB)	Z_1	Z_2	$Z_3/2$	Z_4	R
-35.00	51.58	5.35	255.44	52.97	54.65
50.00	50.00				
-34.00	51.77	5.68	242.96	53.32	55.20
50.00	50.00				
-33.00	51.97	6.03	231.29	53.70	55.82
50.00	50.00				
-32.00	52.21	6.41	220.37	54.12	56.51
50.00	50.00				
-31.00	52.46	6.81	210.17	54.59	57.28
50.00	50.00				
-30.00	52.75	7.25	200.66	55.12	58.15
50.00	50.00				
-29.00	53.00	7.71	191.81	55.71	59.13
50.00	50.00				
-28.00	53.44	8.20	183.58	56.36	60.23
50.00	50.00				
-27.00	53.84	8.73	175.95	57.08	61.47
50.00	50.00				
-26.00	54.29	9.30	168.90	57.89	62.86
50.00	50.00				
-25.00	54.79	9.92	162.40	58.79	64.43
50.00	50.00				
-24.00	55.36	10.58	156.44	59.80	66.20
50.00	50.00				
-23.00	55.99	11.29	151.00	60.92	68.21
50.00	50.00				
-22.00	56.69	12.06	146.06	62.17	70.49
50.00	50.00				
-21.00	57.49	12.90	141.62	63.56	73.08
50.00	50.00				
-20.00	58.38	13.81	137.67	65.13	76.04
50.00	50.00				
-19.00	59.39	14.80	134.20	66.88	79.43
50.00	50.00				
-18.00	60.52	15.88	131.21	68.84	83.33
50.00	50.00				
-17.00	61.80	17.06	128.72	71.05	87.82
50.00	50.00				
-16.00	63.26	18.35	126.72	73.54	93.04
50.00	50.00				
-15.00	64.91	19.79	125.25	76.36	99.12
50.00	50.00				
-14.00	66.80	21.37	124.32	79.55	106.27
50.00	50.00				
-13.00	68.96	23.14	123.98	83.20	114.75
50.00	50.00				
-12.00	71.46	25.12	124.29	87.37	124.88
50.00	50.00				
-11.00	74.37	27.36	125.32	92.19	137.14
50.00	50.00				
-10.00	77.79	29.91	127.18	97.81	152.16
50.00	50.00				

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Insertion Loss:			
Spec. Max.	0.5 dB	1.0 dB	2.0 dB
Typ. Max.	0.2 dB	0.5 dB	0.7 dB
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(Degrees at Max. Freq.)			
Phase Shift			
(Degrees/GHz/Shaft Turn)	1.48		

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values satisfy the symmetry equations.

The element values of Table 1 were used to design a 0.2 GHz to 2 GHz bandwidth balun and Figure 7 shows good agreement between the theoretical and measured frequency responses of the experimental balun.

CONCLUSIONS

A set of graphs has been presented which enables fourth order Marchand baluns with Tchebysheff passband responses to be designed rapidly for a large range of load resistances and bandwidth ratios, without the designer having to resort to the numerically complicated synthesis procedure.

ACKNOWLEDGEMENT

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J. H. Cloete (M'77) was born in Clocolan, South Africa, on June 29, 1945. He received the B.Sc., B. Eng. degrees in electrical engineering from the University of Stellenbosch, Stellenbosch South Africa in 1968. Mr. Cloete received the M. Sc. degree in electrical engineering from the University of California, Berkeley, in 1972. Since 1970, he has been with the South African Council for Scientific and Industrial Research, Pretoria, S.A. working on microwave components and systems. From August 1972 to March 1973, he worked at Scientific-Atlanta, Inc., on the design and manufacture of a multi-octave shaped beam reflector antenna. From March 1977 to February 1978, he was on leave at the University of Stellenbosch working on microwave linear-phase filters.



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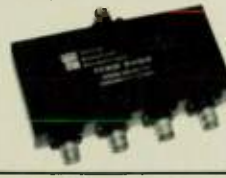
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 Phase Bal. 3°
 Amp. Bal. 0.3dB
 Input Power 2.5w

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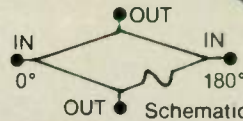
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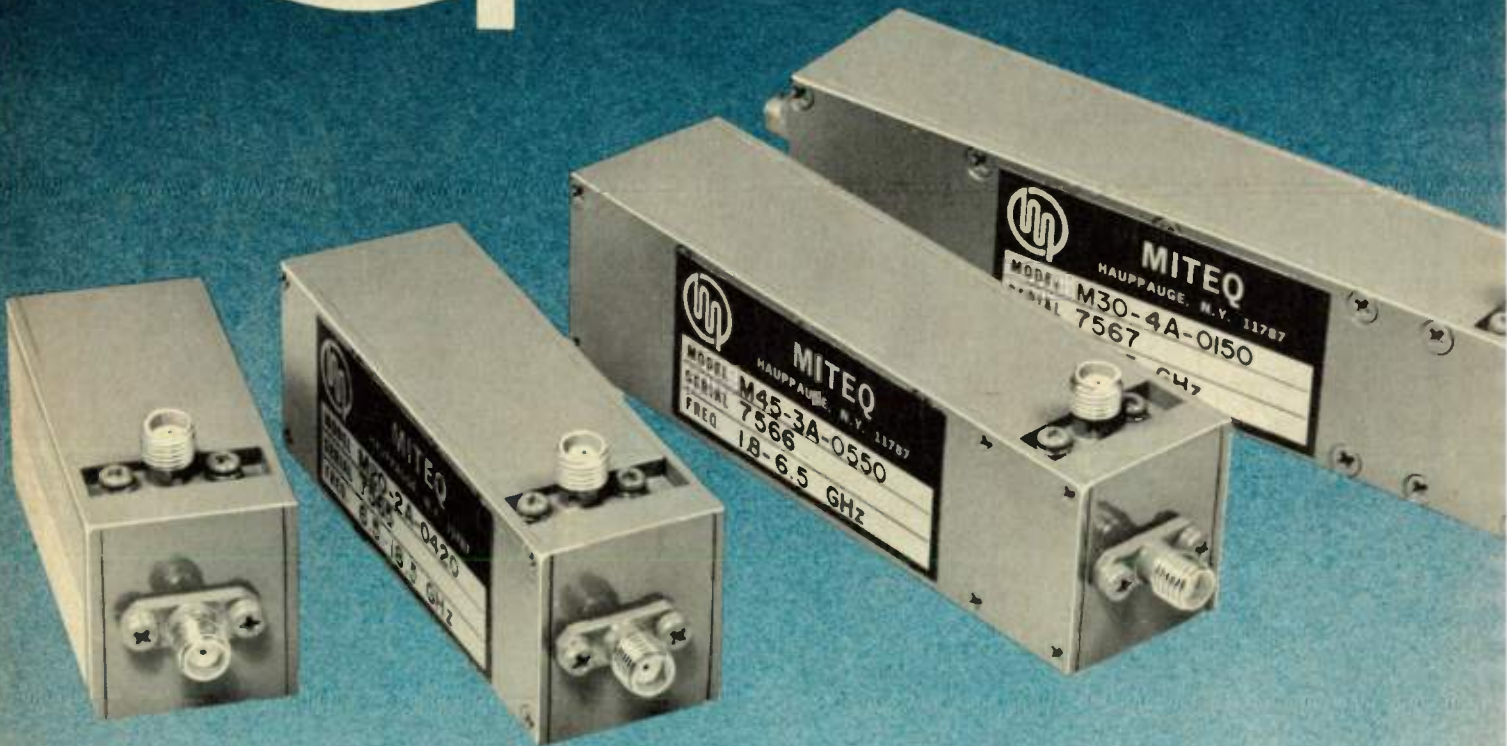


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1-150 MHz IF AMPLIFIER							
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M45-2A-0115	1.8-6.5	22	7.0	8.0	0	40	6
M35-2A-0115	3-10	22	7.5	8.5	0	40	6
M30-2A-0115	4-15	22	7.5	8.5	0	40	6
M20-2A-0115	6.5-18.5	22	8.0	9.0	0	40	6
M2C-2A-0115	2-18.5	22	8.0	9.0	0	40	6
1-500 MHz IF AMPLIFIER							
MLO-2A-0150	1-2	22	7.0	8.0	8	55	2
M45-2A-0150	1.8-6.5	22	7.5	8.5	8	55	2
M35-2A-0150	3-10	22	7.5	8.5	8	55	6
M30-2A-0150	4-15	22	8.0	9.0	8	55	6
M20-2A-0150	6.5-18.5	22	8.5	9.5	8	55	6
M2C-2A-0150	2-18.5	22	8.5	9.5	8	55	6
5-1000 MHz IF AMPLIFIER							
M45-2A-000510	2-6	17	8.5	9.5	0	35	6
M35-2A-000510	3-10	17	9.5	10.5	0	35	6
M30-2A-000510	4-15	16	9.5	10.5	0	35	6
M20-2A-000510	6.5-18.5	16	10.5	11.5	0	35	6
400-2000 MHz IF AMPLIFIER							
M35-2A-0420	3-10	14	8.0	9.0	5	35	6
M30-2A-0420	4-15	14	8.5	9.5	5	35	6
M20-2A-0420	6.5-18.5	14	9.0	10.0	5	35	6

MIXERS

Model	RF & LO Band (GHz)	IF Band (GHz)	Conv. Loss (dB)		Min Isolation	
			Typ.	Max.	LO/RF	LO/IF
OCTAVE BAND DESIGNS						
MLO	1-2	0-0.5	5.0	6.0	25	15
MSO	2-4	0-1.0	5.0	6.0	25	20
MCO	4-8	0-2.0	6.0	7.0	25	20
MXO	8-12.5	0-3.0	6.5	7.5	25	20
MKO	12-18.5	0-3.0	6.6	7.5	20	15
M2D	4-8	8-16	7.5	8.5	25	25
M4D	2-4	4-8	7.5	8.5	25	25
WIDEBAND MULTIOCTAVE DESIGNS						
M45	1.8-6.5	0-1.0	6.0	7.0	25	20
M35	3-10	0-2.0	6.0	7.0	20	20
M30	4-15	0-2.5	6.5	7.5	20	15
M20	6.5-18.5	0-3.0	7.0	8.0	20	15
M2C	2-18.5	0-6	7.0	8.0	20	15
M2C	3-12	0-6	6.5	7.5	25	25

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Arcsprayed Metal Coating For EMI/RFI Shielding

MERLE L. THORPE
TAFE Metallisation, Inc.
Bow (Concord), NH

INTRODUCTION

Stray signals of EMI (electromagnetic interference) and RFI generated by static sparks or outside signals can effect many types of sensitive and critical electronic equipment such as computers, medical equipment, navigational instruments, and process controls. EMI/RFI from auto ignition can produce radio static; likewise, EMI/RFI can result in transit errors in computer operations and cause pacemakers to fault. Electromagnetic interference has become a serious problem, even to the laymen, with the advent of hand tools, household appliances, CB's, and the like.

Conventional shielding methods include foil, tape, screening, plating, vacuum metallizing, conductive paint coatings and metal spraying. The Arcspray process discussed here is a significant improvement over conventional metal spraying techniques. Prior to the development of this process, metal spraying required the use of flames which produce more oxide in the coatings, lower

bond strengths, and significant substrate heating. The Arcspray process is a novel system for melting and spraying any conductive material available in a wire form. It minimizes or eliminates the problems previously associated with flame spraying.

Metal spraying equipment which utilizes a heat source, compressed air and metal wire feedstock has been in existence for a long time. The process involves melting wire, atomizing it and propelling the molten metal particles to the surface to be coated. In the case of EMI shielding, the most desirable metal is zinc because of cost, physical properties, ease of application, minimum toxicology problems, inertness, stability and superior film properties once sprayed. Prior to the advent of Arcspray, metal spraying systems for the application of zinc required the use of an oxygen fuel combustion flame to melt zinc wire (see Figure 1).

A single 11 gauge or 1/8" wire is surrounded by the flame and about 2 kW is transferred to the

wire. The remainder of the energy exists as a hot gas stream impinging on the substrate. This creates problems with the properties of the coating and more importantly, overheats and warps the substrate. The system, by comparison, (see Figure 2) can melt zinc at the same rate with only 2 kW with essentially all the energy used to melt the zinc. Cool compressed air then atomizes the molten surfaces on the melting wires and produces a stream of

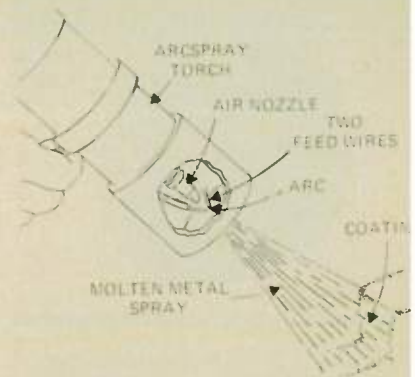


Fig. 2 Arcspray metallizing system.

fine droplets (5-100 micron) suspended in a cool air stream. This permits 1-10 mil coatings to be built up on plastic substrates at rates up to 312 ft.²/hr. without noticeable heating—the substrate temperatures never rising above 100-110°F even though zinc melts at 770°F (see Figure 3).

HOW THE SYSTEM WORKS

The system is, simple and easy to use. Two electrically isolated wires—the coating metal—are fed into the gun so that only the ends of the wires come into contact. An electric arc melts the

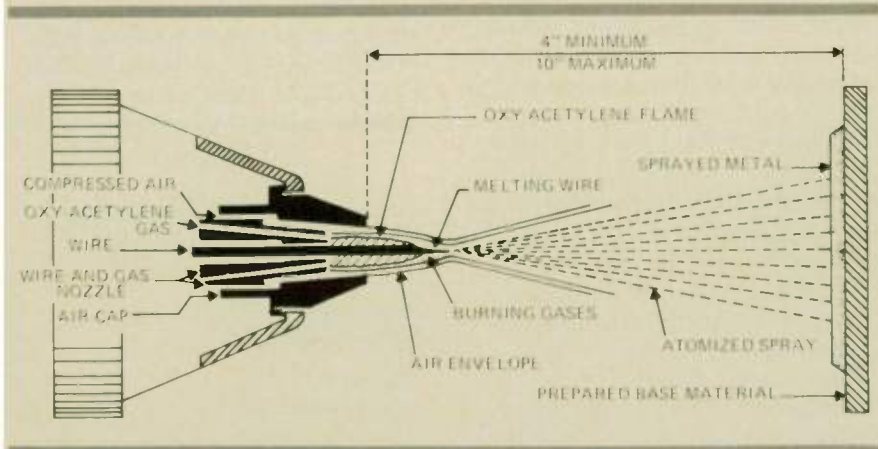


Fig. 1 Oxy fuel combustion metallizing gun.

TABLE I

ARCSPRAYED ZINC COATING PROPERTIES

Tensile strength of zinc coating material	13,200 psi
Density	6.42 gm/cc
Density percent of theoretical	90 percent
Coating weight	0.0334 lbs/ft ² /mil
Conductivity (2 mil coating)	20 milliohms/square
Hardness (Knoop ₁₀₀)	142
Surface finish	55-100 rms
Melting point	770°F
Zinc wire purity (min)	99.99 percent
Life of coating	unlimited
Bond strength	See Table III

TABLE II

SPRAY CONDITIONS

Wire size	1/16" (1.6 mm)
dc Amperes	50-125
dc Volts	20
Air flow	35 SCFM
Air pressure	90 psig
Standoff	6-12 inches
Area coverage (50-125 amps)	
1 mil	124-312 ft ² /hr.
4 mil	31-78 ft ² /hr.



Fig. 3 Zinc with 770°F melting point can be sprayed on any substrate without overheating.



Fig. 4 TAFE Arcspray zinc spraying system with bulk wire dispensers.

ends and the molten material is blown off as fine particles by an air atomizing jet to form a dense, well-bonded, instantly dry metallic coating on the part being sprayed.

The spray gun is similar in size and appearance to those used for paint spraying with the wire feedstock fed automatically. Although the deposits are applied in a molten state, a significant characteristic is that the substrate undergoes only a small temperature rise. The system consists of a spray gun, constant potential power supply, simple control unit and wire dispensing system as shown in Figure 4. Spray rate is controlled by one valve on the spray gun. Input requirements are only 5 kVA and 35 CFM compressed air.

PROPERTIES OF THE COATING

It should be noted that the coating is all metal using 99.99 percent zinc as feedstock, and as long as there is visual coverage, adequate shielding is achieved for most purposes. Zinc coatings have been applied for corrosion protection for many years and considerable technical data is available relative to the life of the zinc coating.

Zinc coatings (unlike paints) have demonstrated a life expectancy of over 40 years in rural outdoor applications. In the case of shielding applications where the environment is even less severe, one could expect unlimited life of the zinc coating. In addition,

if the zinc coating ages, this new surface is also conductive. This eliminates any concern relative to performance of zinc coatings in any foreseeable atmosphere related to electrical and electronic work.

The tables shown above present coating properties and application data. Similar properties are achieved (except for coating finish and visual cosmetic appearance) at 65 psi atomizing but 90 psi is recommended.

The bond of the coating to the plastic substrate is a mechanical one. Thus, the substrate must be either naturally rough as produced or roughened by some artificial procedure. The latter is normally accomplished in production by light grit blasting of the surface of smooth materials such as Lexan, Noryl and smooth structural foams. On a surface blasted with 80 mesh aluminum oxide at 80 psi (standard production procedure) the following bond strengths to the substrate plastic have been achieved

TABLE III

TENSILE BOND STRENGTHS* (psi) ARCSPRAYED ZINC

Noryl	350
Polycarbonate	380
Structural Foam	425
ABS	240
Phenolic	450
Glass Filled Nylon	1050

*Coating pulled at right angles to plastic substrate.

(continued on page 136)

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JUST LOOK AT THESE SPECIFICATIONS

SPECIFICATIONS	Passband (GHz)	Insertion Loss (dB Max)	VSWR (Max)	Stopbands (GHz) 60 dB Min	Power Rating (Watts, average)	Weight (Oz)	Size (Inches) Excluding connectors
FCL-712	1.0—2.0	0.75	1.5	DC—0.95 2.15—6.0	100	1.5	.5 x 1.35 x 2.5
FCL-714	1.0—4.0	0.75	1.5	DC—0.95 4.3—13.0	100	1.25	.5 x 1.09 x 2.29
FCL-718	1.0—8.0	0.75	1.6	DC—0.95 8.6—20.0	50	1.25	.5 x 1.09 x 2.22
FCL-724	2.0—4.0	0.75	1.5	DC—1.9 4.3—13.0	50	1.0	.375 x .9 x 1.85
FCL-728	2.0—8.0	0.75	1.5	DC—1.9 8.6—20.0	50	1.0	.375 x .9 x 1.78
FCL-7212	2.0—12.0	0.75	1.6	DC—1.9 12.9—20.0	50	1.0	.375 x .9 x 1.72
FCL-748	4.0—8.0	0.75	1.5	DC—3.8 8.6—20.0	50	1.0	.375 x .69 x 1.61
FCL-7412	4.0—12.0	0.75	1.5	DC—3.8 12.9—20.0	50	1.0	.375 x .69 x 1.57
FCL-7418	4.0—18.0	0.75	1.6	DC—3.8 19.3—26.5	50	1.0	.375 x .69 x 1.56
FCL-7812	8.0—12.0	0.75	1.5	DC—7.6 12.0—20.0	50	0.75	.375 x .82 x 1.45
FCL-7818	8.0—18.0	0.75	1.6	DC—7.6 19.3—26.5	50	0.75	.375 x .82 x 1.42
FCL-71218	12.0—18.0	0.75	1.6	DC—11.4 19.3—26.5	50	0.75	.375 x .68 x 1.28

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INTERFERENCE SIGNAL ATTENUATION

Normally 2 to 5 mils of zinc spray is adequate to achieve 20 milliohms or less per square surface resistivity. Typically, in shielding, 1-10 ohms per square is considered a good number; 500 ohms per square is adequate for some shielding and electrostatic dissipation, and 2000 ohms per square is effective only for electrostatic discharge, not shielding.

Figure 5 shows the attenuation of a typical 4 mil zinc coating on Lexan polycarbonate. All tests discussed were made in the same apparatus using 4 x 5 panels 1/8" thick. The general conclusions drawn are that for most shielding and all static discharge applications a 2-4 mil coating of zinc is more than adequate. In special cases, additional material or an overlay of copper can improve attenuation performance in certain narrow bands.

A variety of zinc coating thicknesses were tested from 2-10 mils and all exhibited similar characteristics in the 0-400 and 770-1000 MHz range. At 400-770 MHz thicker coatings exhibited greater attenuation. For example, 10 mils gave 20 dB more attenuation than 4 mils at 600 MHz. Overlays of 1, 3 and 5 mils of copper on 4 mils of zinc were also evaluated. As an approximation, all the copper coatings increased attenuation in the 200-450 MHz range (about 20 dB) and a similar improvement was

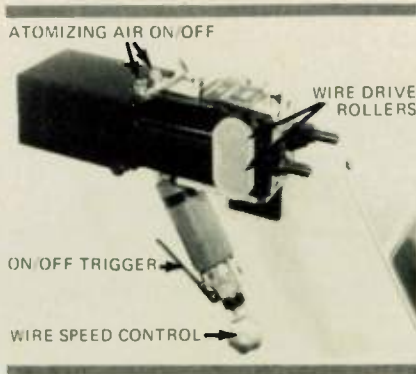


Fig. 6 Arcspray shielding gun.

achieved in the 850-950 MHz range.

As a generalization, 1-2 mils of Arcsprayed zinc appears adequate for static discharge shielding. However, 3-4 mils is usually specified since coating application is by hand and visual coverage results in this thickness when the coating is applied in production.

In the case where waveforms must be attenuated entering or leaving an enclosure, the frequency spectrum should be known and tests run on the specific enclosure. However, it would appear from these measurements that Arcsprayed zinc gives adequate shielding except for very special high energy applications.

EQUIPMENT FEATURES AND PROCESS ADVANTAGES

The Arcspray shielding system is shown in Figure 4. The bulk wire source (drum) is used for volume applications to reduce wire cost through direct mill shipments, elimination of spooling costs, and reduced down-

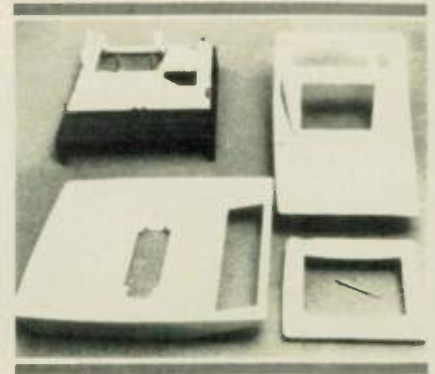


Fig. 7 Arcspray zinc shielded plastic housings.

time for wire pack changes. Twelve hundred pounds of wire is attached to the gun at each loading. This allows continuous spraying for approximately 48 hours. The barrels of zinc are capped with a dispensing system which eliminates tangles, wipes the wire clean of any residual debris, and prevents dust from collecting on the wire prior to feeding. Wire is fed directly into lightweight conduits which are attached to the manual or automated spray gun. The distance from the zinc source to the gun is limited to ten feet of conduit to minimize friction.

The spray gun is shown schematically in Figures 2 and 6 and incorporates a 1/3 horse-power, simple air motor drive, which is positively connected to all four drive rolls. Wire drive rolls are external for easy cleaning and inspection. The wire is pulled from the source to the gun, and pushed an additional four inches through metal tubes which serve as electrical contact for the wires and a guide so that the wires intersect precisely for consistent arcing and efficient atomization. Guide tubes are made in two parts — a permanent tube and an inexpensive replacement tip which can be changed without effecting alignment. Typically such tips are replaced between 200 and 1,000 pounds of sprayed wire. A trigger control lever is located on the spray gun handle to instantly turn the spray on and off.

The general characteristics of this technique which should be considered in any design application include:

- One hundred percent metal coating which gives excellent

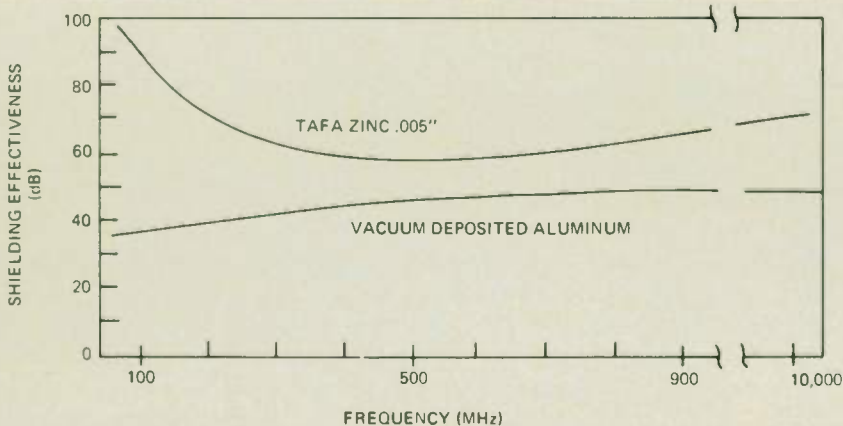


Fig. 5 EMI shielding effectiveness vs. frequency.

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mating surface conductivity, as well as strength and integrity, permits rough handling and attachment of screws and star washers.

- Low temperature application with no after processing or parts heating.
- Instantly dry-part can be handled and used immediately after spraying.
- Easy masking of areas which do not require a coating.
- Instant startup and shutdown and no cleanup of guns, pots and hoses as with paints.
- Unlimited coating life with no deterioration.
- Requires inexperienced operator, visual coverage adequate.
- Application to all types of plastics.
- Low initial investment and coating costs.

ECONOMICS

Estimated costs of the application of the Arcspray zinc coating depends on the number of parts to be sprayed, their complexity, and size. It is difficult to arrive at a generalized cost number.

A survey has indicated the lowest vendor price for a simple Arcspray shielded part (which included no masking) was in the range of \$1.25 per square foot and upward depending on complexity and volume. Depending on the complexity of the part and the number of bosses, screw holes, etc., the survey indicated masking costs could add from 10 percent to 100 percent. Also, one should note, that one coat with a metal is sufficient to produce the shielding results required.

Merle L. Thorpe has been President of TAFE since 1961. He has been an innovator in high temperature technology for the past 25 years, having founded and developed numerous companies in that field. Since 1970, his company has been active in the sprayed metal tooling field. He holds over 30 patents in the field, is a member of the American Welding Society Thermal Spray Executive Committee, and is a member of numerous technical societies including the Society of Plastics Engineers. Mr. Thorpe holds an A.B. and M.S. in Engineering Science from Dartmouth College. In addition to his industrial activity, he has been a Professor and Assistant Dean at Dartmouth.

Cover Story

K-Band Fundamental YIG-Tuned Oscillator



AVANTEK, INC.

An 18-26.5 GHz YIG tuned oscillator integrated with a multi-stage buffer amplifier extends the range in which fundamental transistor oscillator sources may be applied and their inherent advantages enjoyed.

The Avantek AV-71826 employs 15 μm gate width, submicron gate length GaAs FET's in its fundamental oscillator and buffer amplifier stages to deliver a minimum of 20 mW over the full 18-26.5 GHz range. Peripheral characteristics offer significant benefits when compared to available bulk diode and multiplied bipolar transistor sources in that frequency band.

Over the 18-26.5 GHz band, maximum output power variation is ± 3.0 dB; pulling at 12 dB return loss is 0.5 MHz, typical; frequency drift over the rated case temperature range of 0° to 65°C is 60 MHz maximum; spurious output is 60 dBc, minimum and, with all harmonics above 36 GHz, filtering for harmonics poses no particular problem.

The main tuning port characteristics of the AV-71826 include a typical current sensitivity of 27 MHz/mA, 3 dB bandwidth of 5 kHz typical and typical linearity of 0.1%. Its FM port sensitivity is typically 450 kHz/mA, its typical 3 dB bandwidth is 400 kHz and its maximum deviation at 400 kHz is 90 MHz.

With a dc drive of 150 mA maximum at 15 volts, the AV-71826 bias power requirement is well below those of bulk diode designs in its frequency range.

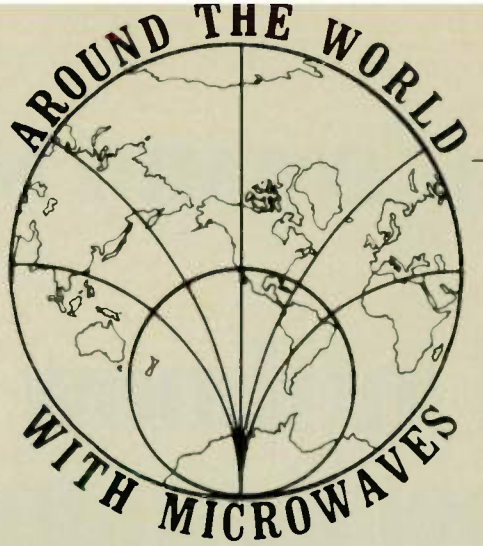
The new oscillator is well suited to instrument and EW system applications. The similarities of its operating characteristics and power requirements to those of its lower frequency predecessors make it a particularly attractive option for extending the range of existing designs and for new broadband equipment which must cover the 18-26.5 GHz band.

Complete specifications are listed in Table I.

TABLE I
AV-71826
Oscillator Specifications
(At 25°C Case Temperature)

Frequency Range	18 to 26.5 GHz
Power Output (50 ohm load), Min.20 mW (+13 dBm)
Power Output Variation, Max.	±3.0 dB
Operating Case Temperature Range.	0 to 65°C
Frequency Drift over Operating	
Temperature, Max.	60 MHz
Pulling Figure (12 dB return loss), Typ.	0.5 MHz
Pushing Figure, Typ. +15 V Supply.	1 MHz/V
Magnetic Susceptibility, Typ. @ 60 Hz	50 kHz/Gauss
Spurious Output (below carrier), Min.	60 dB
Main Tuning Port Characteristics	
Sensitivity, Typ.	27 MHz/mA
Bandwidth (3 dB), Typ.	5 kHz
Linearity, Typical	0.1%
Hysteresis, Typ.	12 MHz
Input Impedance, Typ. @ 1 kHz	12 Ω in series with 85 mH
FM Port Characteristics	
Sensitivity, Typ.	450 kHz/mA
Bandwidth (3 dB), Typ	400 kHz
Deviation @ 400 kHz Rate, Max.	90 MHz
Input Impedance, Typ.	1 ohm in series with 1.25 μH
DC Circuit Power, Max.	+15V @ 150 mA Max.
YIG Heater Power	20 to 28 VDC, 2 W max. @ 25°C 3 W max. @ 0°C
Weight, Nominal	17 oz.
Outline Dimension	YIG M3 with no -5V connection and center to connector spacing of .600

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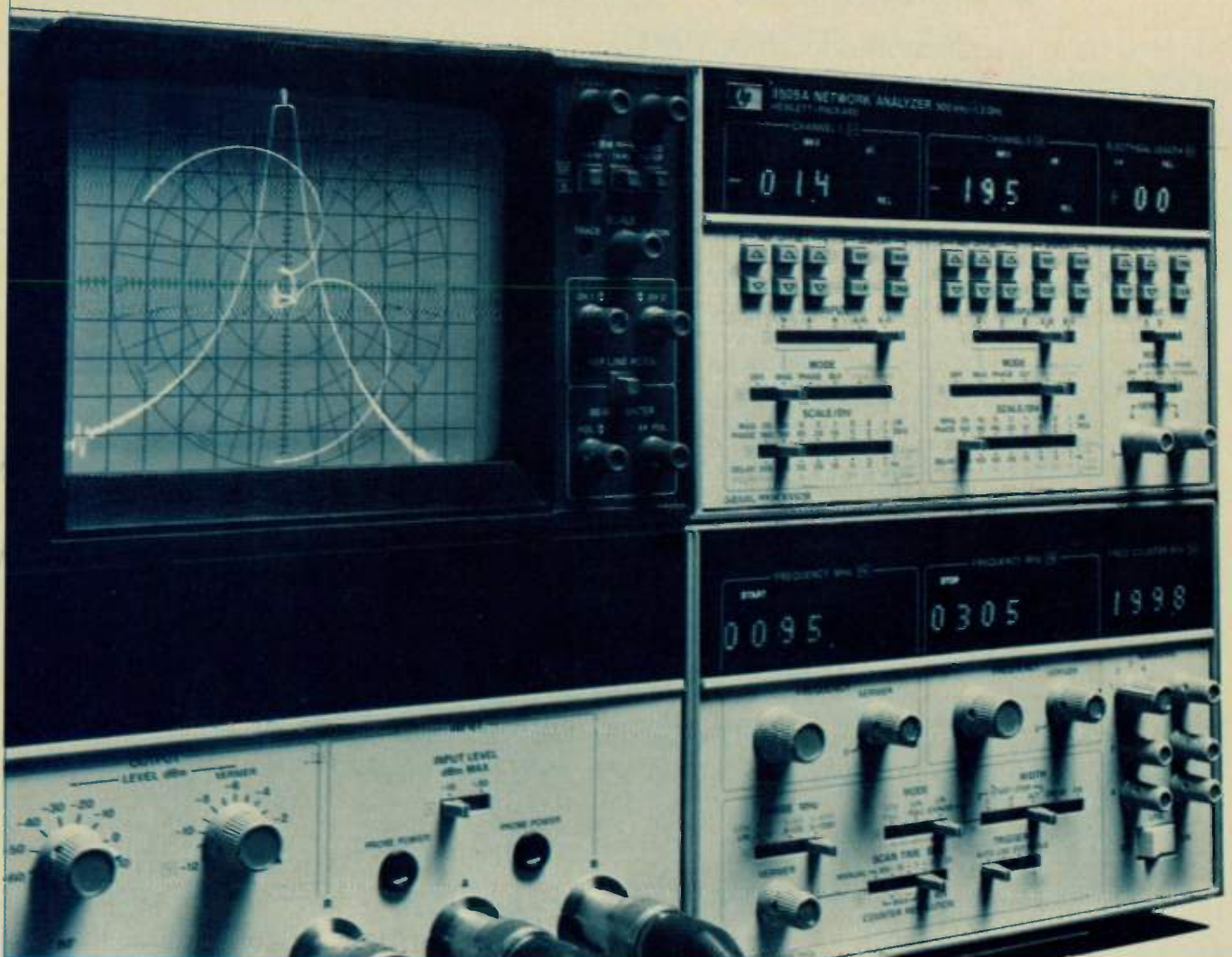
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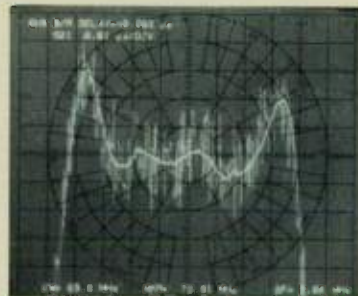
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We've only touched on the highlights of the 8505A's performance and capabilities here. For complete data, contact your nearby HP field sales office, or write 1507 Page Mill Road, Palo Alto, CA 94304.

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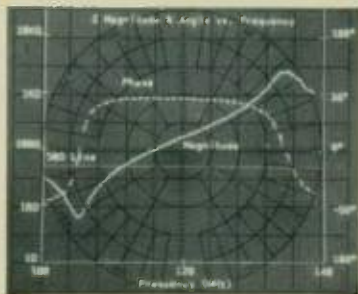
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Cucamonga, CA

Precise measurements of components employing SMA coaxial connectors have been hampered by the lack of an efficient reference plane short circuit device. The characteristics of the widely used "buried contact" design are far from ideal beginning at 3 GHz and they deteriorate rapidly and continuously as frequency increases.

Designed to eliminate these deficiencies, Maury Microwaves' Model 360D Precision Female Fixed Short employs a large slotted female contact whose outside diameter equals the outside diameter of the mating plug and the entire face of the slotted piece serves as the shorting plane. Upon mating, the outer conductor of the male connector bottoms on the face of the slotted contact. When the coupling is tightened, the slotted contact is compressed around the male pin to complete the short circuit.

Figure 1 illustrates the characteristics of two SMA "buried contact" shorts and one Model 360D short measured against a standard 0.5 cm. offset fixed short. Tabulated amplitude and phase data for one sample of each design are shown in Table 1.

The Model 360D mates with APC 3.5, SMA, MPC3 and MPC4 connectors and is designed for use to 40 GHz. Minimum reflection coefficient through 18 GHz is 0.98; the company will supply specification data at frequencies between 18 and 40 GHz.

The original design for the Model 360D was developed by R. Stewart of Hewlett-Packard.

Circle 132 on Reader Service Card.

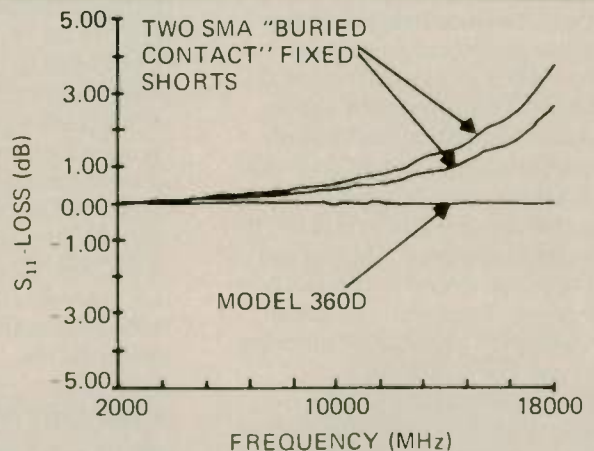


Fig. 1 Reference phase short circuit return loss.

TABLE I

FREQUENCY MHz	MODEL 360D RETURN LOSS-IN S_{11}		SMA "BURIED CONTACT" SHORT RETURN LOSS-IN S_{11}	
	dB	Ang	dB	Ang
2000	-0.00	-180.0	.07	178.4
10000	-0.00	-179.9	.20	169.1
18000	-0.01	-179.9	4.46	123.0

Microwave Products

MICROPROCESSOR-BASED COUNTER EXTENDS CAPABILITY TO 110 GHz

DUAL FREQUENCY LOCKBOX

Instrumentation

PEAK RF POWER METER



The PK-38 is a peak RF power meter designed to indicate the peak value in dBm, of an RF pulse, at least 10 μ s wide in any time frame at any repetition rate. This time rep-rate range also includes CW. The 3-digit LED display indicates peak power in a 20 dBm range. Offset adjustments have a range of \pm 99.9 dBm. Frequency range is 1 GHz to 13.5 GHz. The meter can be calibrated from either a peak or CW source. Price: \$2,200, per unit. Del: 90 days. ITM Systems, Inc., Sunnyvale, CA. Walter Berg, (408) 745-6247. **Circle 163.**

PHASE-CONTINUOUS FREQUENCY SYNTHESIZER



Model 6757 is a frequency synthesizer which mixes the output of a 0-5 MHz direct, digital synthesizer with a 67.5 MHz signal to cover a minimum tuning range of 650 kHz centered at 70 MHz with a step size of 0.298 Hz. Both mixing signals are derived from an external 10 MHz reference. The absence of phase-locked loops yields sub-microsecond switching; frequency changes are phase-continuous and phase is digitally-controllable to within 0.1 degree. Output power is +10 dBm, spurious outputs are -50 dBc, and single sideband noise is -95 dBc/Hz from 75 Hz to 3 MHz from the carrier. Inputs are +15 and -5.2 V and commands are ECL-compatible. Meets requirements of MIL-E-5400R, Class II environment and MIL-P-11268, MIL-S-19500 and MIL-M-38510. Size: 70 cu. in. Weight: 4 lbs. Zeta Labs, Inc., Santa Clara, CA. (408) 727-6001. **Circle 180.**



Automatic frequency measurement from 10 Hz to 110 GHz is available from the Model 548A microprocessor-based microwave counter equipped with Option 06 Frequency Extension Module, Model 590 Frequency Extension Cable and a series of Remote Sensors. A readout up to 12 digits at all frequencies is displayed in approximately 1 second. Exclusive of options, the Model 548A range extends to 26.5 GHz. The full option set provides 26.5-40, 40-60, 60-90 and 90-110 GHz measurement bands. Sensor sensitivity is -25 dBm typical, with typical maximum input at +5 dBm, and damage level at +10 dBm. Standard features include multiple signal discrimination, frequency offsets, internal diagnostics and signature analysis and front-panel keyboard control. Price: Model 548A, \$6,300, the complete option set, \$1,550. Avail: 16 weeks ARO. EIP Microwave, Inc., San Jose, CA. Howard Lurie, (408) 946-5700. **Circle 177.**

FULLY PROGRAMMABLE UNIVERSAL COUNTER

Option 040 full programmability for the HP Model 5335A Universal Counter provides complete front panel programming of the instrument from an interface bus (IEEE-488). Complete control over all of the counter's front panel measurement and input signal conditioning controls and full remote programmability for single-shot time interval measurements is provided. The option also allows the remote matching of the counter's input to various signal level and input matching situations. Option 040 adds \$700 to the base price of the Model 5335A counter. Hewlett-Packard, Palo Alto, CA. (415) 857-1501. **Circle 178.**

LOGARITHMIC AMPLIFIER/METER

Model LN-70 is a logarithmic amplifier/meter with a dynamic range of 70 dB, +20 to -50 dBm. An offset control on the front panel enables the instrument to measure either gain or loss directly to 0.1 dB resolution. Measurements are taken directly from a digital readout; powering is from 115/230 Vac, 50/60 Hz, with a power consumption of less than 15 W. Price: \$595. Texscan Corporation, Indianapolis, IN. Raleigh B. Stelle, III, (317) 357-8781. **Circle 215.**



The MOSP 3000 Series can be used to phase lock two RF sources at a constant offset frequency or to stabilize a single source to the stability of an internal quartz crystal oscillator. Reference to an external 5 MHz standard is optionally available. 11 standard offset frequencies are available and others may be predetermined. The instrument will phase lock both solid state and tube source up to 18 GHz, special models are available to 40 GHz. Price: \$4,500. Microwave Systems, Inc., East Syracuse, NY. (315) 437-9951. **Circle 179.**

BUS INTERFACE UNIT TIES RF WATTMETERS TO GPIB



An IEEE-488 Bus Interface Unit (Model 4380-488) provides connection of an RF wattmeter line to a General Purpose Instrumentation Bus (GPIB) for automatic measurement and recording of data. Digital THRULINETM Wattmeters equipped with a bus-compatible output can now become a talker and a listener via GPIB control. Line of digital RF Power AnalystTM Wattmeters operate from .5 to 2300 MHz at 100 mW to 250 kW, depending on model and plug-in elements. Model 4380-488 price: \$975. Del: 4 wks. ARO. Bird Electronic Corporation, Cleveland, OH. H. H. Heller, (216) 248-1200. **Circle 216.**



PHASE MATCHED CABLE ASSEMBLIES...

As You Like Them

When your requirements call for phased matched cable assemblies, Uniform Tubes has your answer. U.T. can provide complete assemblies, in any quantity desired, individually tested on an automatic network analyzer, and at modest cost. Print outs of graphs or hard data are available for each assembly.

Phase matching with the automatic network analyzer is also available for the U.T. line of "In-A-Cable" low-pass and band-pass filters. The analyzer can also save you money in testing standard cable assemblies requiring insertion loss and VSWR data, and coaxial delay lines requiring exact delay measurements.

Let us show you how you profit when you deal with Uniform Tubes high quality MicroCoax® cable assemblies, "In-A-Cable" filters, and coaxial delay lines. Contact us by phone or send for details on phase matched assemblies.

The smaller your package gets, the better we like it.



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COLLEGEVILLE, PA 19426, U.S.A. • PHONE 215/539-0700
TWX 510-660-6107 • Telex 84 6428

(from page 143) PRODUCTS

CONNECTOR GAGE SETS

GS series connector gage sets are offered for measuring longitudinal dimension deviations in SMA, N and TNC/BNC connectors. Sets contain two gages, one for measuring male and the other for measuring female contact pin location. Gages are portable and self-calibrating and are packaged in their own case. The SMA set includes a gage for measuring the height of the insulator. Size: dial indicator - 2.25" diameter, dial calibrator - 0.001" increments, 100/360°; case dimension - 6" D x 10" W x 3 5/8" H. RLC Electronics, Mt. Kisco, NY. (914) 241-1334. Circle 164.

Systems

200 kW, X-BAND RADAR SUBSYSTEM

A 200 kW, X-Band (8.6-9.4 GHz) radar subsystem, Model VZX-3424, is designed for shipboard applications, particularly moving-target indicator systems. Unit uses a multichannel, crystal-referenced local oscillator (STALO) and an integral transmitter AFC that permits high-speed selectivity of up to 12 separate frequency channels by panel control or TTL signal. Minimum peak output power is 200 kW, pulse width is 0.5 μsec and repetition rates to 1800 Hz can be accommodated. Subsystem includes a solid state modulator, magnetron, magnetron tuner, power supply, control/monitor panel, an RF isolator and cooling fan. Varian Associates, Beverly, MA. (617) 922-6000. Circle 169.

OSCILLOSCOPE MEASUREMENT SYSTEM

HP 1980 is a microprocessor-controlled oscilloscope measurement system for automated measurements. This a complete stand-alone instrument that does not require plug-ins or enhancement to make measurements. The HP-IB (IEEE-488) capability is standard and allows programming of eight complete front panel set-ups. Deflection factors are 2 mV per div. at 100 MHz. Two independent 5 ns sweep speeds for both main and delayed sweeps, main and delay trigger view, a 10 cm by 12 cm CRT with a 10 x 10 division internal graticule, a dual (main and delayed) sweep mode, both delta time and delta voltage measurements, and menu capabilities are all standard. The HP 1980 has automatic signal acquisition capability and selective auto-scope function, as well as continuous calibration and CRT and character readout. Price: \$8,500. Del: from November 1980. Hewlett-Packard Co., Palo Alto, CA. (415) 857-1501. Circle 216.

Materials

ELECTRICALLY CONDUCTIVE PRESSURE SENSITIVE ADHESIVE

A series of low cost, electrically conductive pressure sensitive tapes, ECCOSHIELD® PST-C, are designed for EMI/RFI shielding applications. Product uses a conductive pressure sensitive adhesive applied to a foil backing. Tape has measured attenuation up to 100 dB over a wide frequency range (30 Hz to 500 MHz). Material is offered in ready-to-use 180 ft. rolls in widths to 10 inches. Price: From 5¢ to 15¢ per lineal ft. depending on width. Emerson & Cuming, Dewey & Almy Div., W. R. Grace & Co., Canton, MA. Joe Flaherty, (617) 828-3300. Circle 165.

Hardware

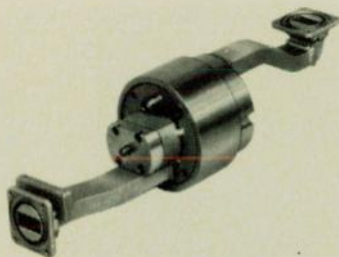
RF ENCLOSURES WITH INTERCHANGEABLE CONNECTORS

A line of RF enclosures with interchangeable SMA, BNC, TNC and N connectors is offered. Users can purchase standard enclosures and additional quantities of connectors (COMPAC SMA, BNC, TNC or N Types). Interchangeable flange mounted "female" connectors have four uniform screw holes matching the screw holes in the enclosure and are usable with standard or custom Compac enclosures. Price: from \$11.75 per unit. Del: From stock. COMPAC, Deer Park, NY. Susan MacDonald, (516) 667-3933. Circle 142.

Components

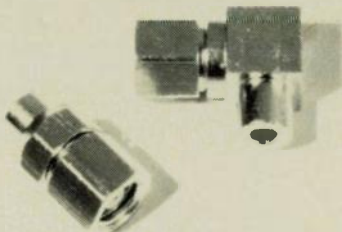
THREE-CHANNEL ROTARY JOINT WAVEGUIDE SPANS 8.5-10 GHz

Model 1379 is a three-channel rotary joint with waveguide channel operating in the 8.5-10 GHz frequency range at 500 kW peak with 100 W average. The center coaxial channel operates at 1-10 GHz with 2 W peak power and 1 W average power. Outer coaxial channel operates from .5 to 1.1 GHz with 6 kW peak power and 20 W average power. Kevlin Manufacturing, Woburn, MA. Ernest Lattanzi, (617) 935-4800. **Circle 212.**



ONE-PIECE, SOLDER TYPE SMC SCREW-ON PLUGS

One-piece, semi-rigid SMC cable plugs (straight Pt. No. 701803 and right-angle Pt. No. 701892 varieties) are designed for .086" cable. These screw-on plugs feature assembly by inserting the cable into the captivated contact and soldering the jacket. Units meet MIL-C-39012B specifications and all metal parts are gold-plated per MIL-G-45204, Type 1, Class 2, Grade C over copper plate per MIL-C-14550, Class 4. Cablewave Systems, Inc., North Haven, CT. Steven Raucci, Jr., (203) 239-3311. **Circle 183.**



BROADBAND PHASE SHIFTER COVERS dc TO 4 GHz

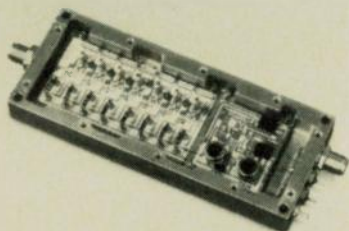
A broadband phase shifter, Model 9428C, operates from dc to 4 GHz and has a linear phase shift of 180° at 1 GHz and 360° at 2 GHz. Insertion loss is less than .5 dB to 2 GHz, 0.75 dB to 4 GHz; SWR is 1.3 max. Size: 12" x 2" x 1"; connectors are SMA. Price: \$475, small qty. Del: 6 wks ARO. Arra, Inc., Bay Shore, NY. Mike Geraci, (516) 231-8400. **Circle 182.**

SATELLITE COMMUNICATIONS SWITCHES

Waveguide satellite communication switches in WR-229 are intended for both low noise amplifier and high power amplifier earth station systems. Typical performance specifications include SWR of 1.04; 0.01 dB insertion loss and 90 dB isolation. Logus Manufacturing Company, Deer Park, NY. (516) 242-5970. **Circle 162.**

LOG IF AMPLIFIERS FOR 2 GHz BAND

An IF-to-log video amplifier series, ICL-5, cover the 600-2000 MHz frequency range. Log amplifier accuracy is typically less than ± 1 dB deviation from ideal log plot at temperatures to 85 C. Options such as hermetically sealing and power supply protection and regulation are offered. Varian Associates, Beverly, MA. (617) 922-6000. **Circle 211.**

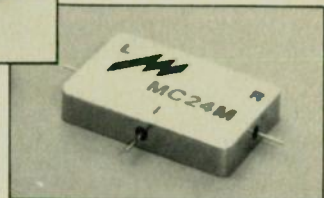
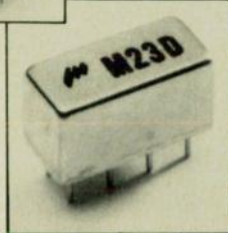
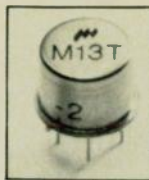


(continued on page 148)

The FIRST FAMILY

DC to 6.4 GHz

Doubled Balanced Mixers



If you are looking for performance and reliability at an affordable price — look to our family of products.

Frequency Range RF, LO (MHz)	IF (MHz)	Conversion Loss (dB)	Isolation L-R L-I (dB) (dB)	Intercept Point (dBm)	Model No.	PRICE QTY. 100
------------------------------------	-------------	-------------------------	-----------------------------------	-----------------------------	--------------	----------------------

Low Level Mixers	(LO input = +7dBm)	TO-8 PACKAGE
0.01 - 65 DC - 65	5.5 55 55	M13T \$ 18.25
0.2 - 200 DC - 200	5.5 55 50	M23T 10.45
0.5 - 500 DC - 500	5.5 45 45	M33T 8.00
5 - 1000 DC - 600	6.0 35 30	M43T 11.64
900 - 1300 DC - 600	6.5 30 30	M43T-1 10.80
10 - 1500 DC - 800	7.0 30 30	M53T 21.50

High Level Mixers	(LO input = +13dBm)	TO-8 PACKAGE
0.01 - 65 DC - 65	5.5 50 50	+16 M16T \$ 19.05
0.2 - 200 DC - 200	5.5 55 50	+17 M26T 16.30
0.5 - 500 DC - 500	5.5 45 50	+17 M36T 12.40
5 - 1000 DC - 600	6.0 35 30	+16 M46T 22.05
900 - 1300 DC - 600	6.5 30 30	+16 M46T-1 21.15
10 - 1500 DC - 800	6.5 30 25	+16 M56T 31.15

Low Level Mixers	(LO input = +7dBm)	DIP PACKAGE
0.01 - 65 DC - 65	5.5 55 55	M13D \$ 18.20
0.2 - 200 DC - 200	5.0 50 40	M23D 10.42
0.5 - 500 DC - 500	6.0 45 35	M33D 8.00

High Level Mixers	(LO input = +13dBm)	DIP PACKAGE
0.01 - 65 DC - 65	5.5 55 55	+17 M16D \$ 19.00
0.2 - 200 DC - 200	5.5 50 40	+17 M26D 17.52
0.5 - 500 DC - 500	6.0 40 35	+17 M36D 13.20

Communications Band Mixers	MAGNUM-PAC™	
(GHz)	(GHz)	
3.7 - 4.2(R) DC - 1.1	4.8 30 25	MC24M \$ 59.60
2.5 - 5.4(L) @+9dBm		MC24P 91.60
5.9 - 6.4(R) DC - 1.1	7.0 20 25	MC36M 78.00
4.8 - 5.3(L) @+13dBm		MC36P 110.00

Magnum Microwave

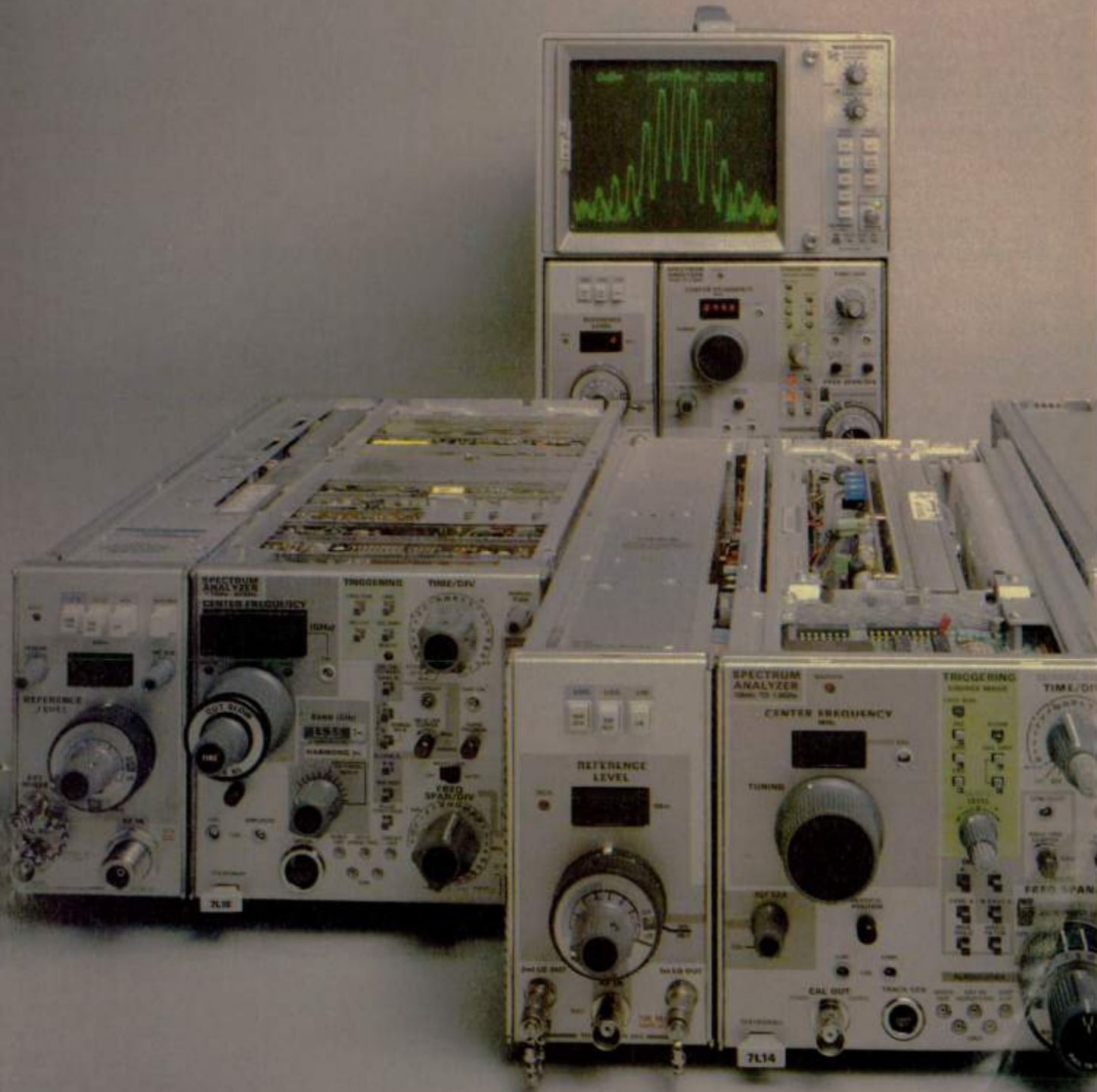


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CIRCLE 190 ON READER SERVICE CARD

TEK 7000 SERIES SPECTRUM ANALYZERS



Now you can plug in an all digital-storage family of spectrum analyzers.

Three models provide laboratory performance with flexibility and versatility.

With the introduction of the new 7L14, there are now three members of the Tektronix family of digital storage plug-in spectrum analyzers. They provide frequency coverage from 20 Hz to 60 GHz. This plug-in concept brings you high performance, versatility and flexibility unmatched by monolithic instruments. At reasonable prices. They're compatible with any Tektronix 7000 Series oscilloscope mainframe, including the new digitizing GPIB

7854 for programmable solutions to complex measurements.

Family characteristics that make spectrum analysis easier.

All three instruments feature digital storage for flicker-free displays that are easy to interpret. This provides averaging and peak detection, accurate waveform comparisons, stores for long periods to measure amplitude changes and frequency drift.

The 7000 Series plug-in family displays alphanumeric readout for referencing and easy documentation. And each spectrum analyzer is protected from up to one watt input levels to save expensive front end repairs caused by inadvertent overloading.

With a 7000 Series mainframe on your bench, you select the spectrum analyzers that fit your requirements. And they interchange quickly with 30 other Tektronix test and measurement plug-ins. Use the powerful mainframe for logic analyzer, oscilloscope and other measurements

New 10 KHz to 1.8 GHz 7L14

completes the 20 Hz to 60 GHz digital storage plug-in team.

Here it is. The new 7L14 for digitally-stored close-in, high resolution measurements from 10 KHz to 1.8 GHz. With 10 Hz residual FM, the 7L14 provides stability and jitter-free 30 Hz resolution displays. Its digital storage can be used to eliminate system errors and provide flat swept RF measurement capability. Digital averaging provides noise reduction which gives 70 dB spurious-free dynamic range. You can check broadband RF networks, filter networks, amplifiers, cables. Measure EMI/RFI and FM, navigation, two-way and

other communications systems.

other communications systems.

At the top of the spectrum you get top performance from the 7L5 Spectrum Analyzer. It provides full amplitude calibration in the 1.5 GHz to 60 GHz range and has 30 Hz resolution to 12 GHz. Displays are sharp, stable and flicker-free. Digital storage and digital signal processing make complex measurements easy with microprocessor aided controls. An automatic preselector insures spurious-free operation, giving easily interpreted displays.

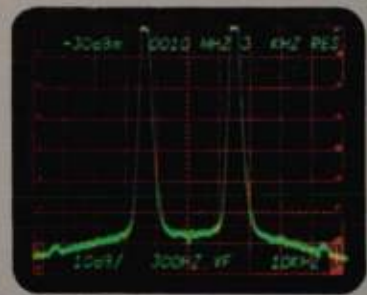
For baseband measurements choose the 7L5 for its precision and convenience in the 20 Hz to 5 MHz range with 10 Hz resolution. For a high performance analyzer, it's unusually easy to operate.

This family works together to make an outstanding value.

You get this laboratory performance and measurement flexibility, prices that point up the value of the Tektronix plug-in concept.

Call your nearest Tektronix Field Office (listed in major city directories for complete details on the 7000 Series lab performance spectrum analyzers. Or call 800-547-1512 for descriptive literature.

Tektronix, Inc., P.O. Box 1700, Beaverton, OR 97075. In Europe: European Marketing Centre, Postbox 827, 1180 AV Amstelveen, The Netherlands.



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Diode Types Tunnel, ZBS, Schottky
RF Conn. Mates w/OSSM Female
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- Our catalog product line of Detectors, Limiters and Diode Switch Modules are available in the standard VHF, UHF and Microwave bands.
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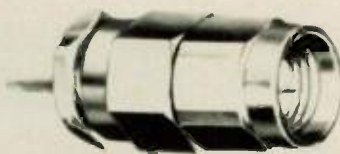
or send for our Product Literature.

CIRCLE 192 ON READER SERVICE CARD

(from page 145) PRODUCTS

SMA PLUG FOR BRAIDED COAXIAL CABLE

Designed for use with RG-174/U, -188/U and -316/U, the Model 50-607-3188 plug features clamp-type cable attachment. Unit is designed for interfacing



these types of braided coaxial to SMA connectors. Size: .750" L. max.; .312" hex body. Constructed of stainless steel and offered in gold-plated or passivated finish. RF Components Div., Sealectro Corp., Mamaroneck, NY. (914) 698-5600. Circle 184.

MULTI-CHANNEL AIRBORNE ROTARY JOINT

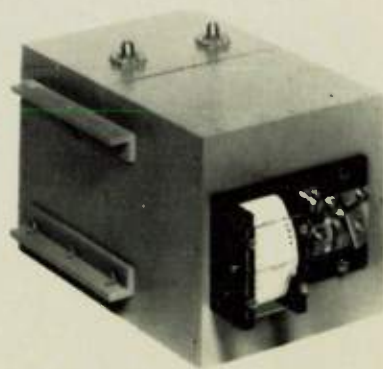
Model DIC-2319 is a 3-channel rotary joint with separate X-band receive and transmit channels in WR 112 waveguide. Power ratings are 10 W and 3 kW CW, respectively; insertion losses are 0.15 dB and 0.2 dB max. The third channel for Q-band is rated at 500 W CW and 0.15 dB max. insertion loss. Other models in C-band, S-band and at 20/40 GHz are available. Diamond Antenna and Microwave Corporation, Winchester, MA. (617) 729-5500. Circle 159.

5.9 - 6.4 GHz UPCONVERTER

The MC36M is an upconverter for the 5.925 - 6.425 GHz band. With an RF input of 1.1 GHz and local oscillator drive at 4.8 - 5.3 GHz, typical conversion loss of the upconverter is 6.3 dB, maximum is 8 dB. At +15 dBm LO drive, the third order intercept is at +18 dBm, typical. MC36P is the SMA connector version of the upconverter. Price: MC36M - \$78, MC36P - \$110 (100 qty.) Magnum Microwave Corp., Sunnyvale, CA. David Fealkoff, (415) 738-0600. Circle 156.

FOUR POLE TUNABLE BANDPASS FILTER

A four pole tunable bandpass filter designed for use in the L band (1350-1850 MHz) is offered. The filter has an insertion loss of 1.5 dB maximum at the center frequency ± 1 MHz and rejects all frequencies beyond the center frequency ± 30 MHz by 70 dB minimum. Power handling capability of 40 W. The device has been constructed to operate in a full military environment, temperature range is -62° to $+85^{\circ}$ C. Calibration is in frequency directly with 1 MHz frequency marks. It may be reset within 100 Hz, a frequency accuracy of 0.008%. Price: \$1575, 1-2 qty. Del: 10-12 wks. K & L Microwave, Inc., Salisbury, MD. Charles Schaub, (301) 749-2424. Circle 161.



VARIABLE POWER HIGH POWER COMBINERS/DIVIDERS

Five series of variable power combiners/dividers designed for communication satellite earth station applications are offered. Waveguide sizes range from WR-187 to WR-75; power ratings range from 10 kW to 2 kW. In the combiner mode, online HPA's has SWR of 1.15 max. in hot switching; insertion loss for all models is 0.2 dB max; isolation between the two HPA inputs is 30 dB min. Continuously or step-variable models provide a total of 40 dB of power adjustment. In the reverse mode, the devices operate as power dividers permitting two antennas to serve a single up-link. Unit price for the WR-137, \$3,200 Del: 60 days. NEICO Microwave Co., Hopkinton, MA. Robert E. Ranslow, (617) 435-6366. Circle 158.

(continued on page 151)

400 - 470 MHz PRESELECTOR

Preselector Model 3779 tunes the 400-470 MHz band with a 3 dB bandwidth of 3 MHz and an insertion loss of 1.5 dB. Selectivity is 30 dB ± 9 MHz from the center frequency. Tuning of the 2-cavity filter is screwdriver controlled, connectors are SMC. Price: \$295, (1-9 qty.). Del: 10 days. Microwave Filter Co. Inc., East Syracuse, NY. Emily Bostick, (315) 437-3953. Circle 157.

SINGLE CHANNEL CIRCULAR WAVEGUIDE ROTARY JOINT

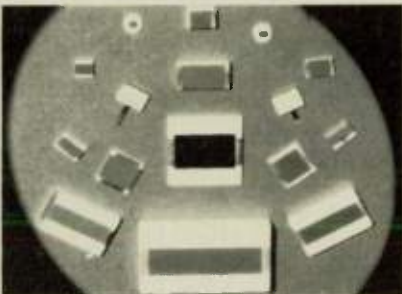
A single channel circular waveguide rotary joint, Model 21015, operates in the TE₁₁ mode at 26.5 to 40 GHz. Power capability is 2 kW CW. Kevlin Manufacturing, Woburn, MA. Ernest Lattanzi, (617) 935-4800. Circle 153.

SPDT COAXIAL SWITCH

Model 50S is a single-pole, double-throw coaxial switch for the .5-500 MHz frequency range. Insertion loss is 1.0 dB maximum, SWR is 1.2 maximum and isolation is 60 dB to 50 MHz, 50 dB to 250 MHz and 40 dB to 500 MHz. Standard control voltage is 26 V at 30 mA. Various packages, control voltages, frequency ranges and connectors are offered; BNC connectors are standard. Price: \$70, 1-9 qty. Del: 2-4 wks. JFW Industries, Inc., Beech Grove, IN. Joetta L. Walker, (317) 783-9875. Circle 152.

POWER CHIP RESISTORS AND TERMINATIONS

A series of power chip resistors and terminations with power handling capabilities from 10 W to 800 W is offered. Resistance ranges from 10-500 ohms; tolerances are 2, 5 and 10 percent. Components meet all test requirements of MIL-STD 202, and come in either thick film or thin metal film deposited on beryllium oxide substrate. Size: 100 x 200 mils to 2,000 x 1,800 mils, depending upon power. Termination, resistor or custom network configurations available. KDI Pyrofilm Corporation, Whippany, NY. Al Arfin, (201) 887-8100. Circle 160.



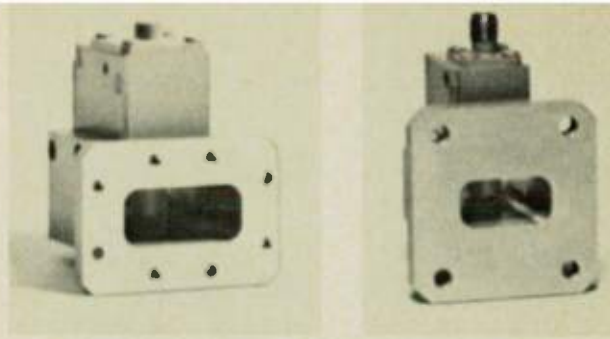
SOLID STATE CLASS A POWER AMPLIFIER

Model 5100L is a solid state Class A power amplifier which covers the 1.5 to 400 MHz frequency range with gain of 50 dB. Linear power output rating is 100 W with more than 200 W available over the 1.5-200 MHz range. Harmonics are -25 dBc at 100 W output, typical 3rd order intermod point is +62 dBm. Model is powered by its own regulated power supply from 115 Vac 50-60 Hz at 22 amperes. Price: \$9150, FOB. Del: Stock to 60 days. Electronic Navigation Industries, Inc., Rochester, NY. L. M. Salmen, (716) 473-6900. Circle 155.

BROADBAND AMPLIFIER MODULES FOR 10 kHz - 1 GHz BAND

A line of solid state linear broadband amplifier modules, covering the 10 kHz to 1 GHz frequency range, are designed for OEM applications. Modules can be tailored for specific power levels, frequency ranges, AM, FM and pulse modulation. Typically, the modules are supplied in EMI shielded enclosures and attached to appropriate heat sinks. User supplies proper dc voltage and forced air cooling, if needed. Push-pull circuit design is used to enhance linearity; modules provide continuous power adjustment. AIL-TECH Electronic Instruments, Eaton Corp., Los Angeles, CA. (213) 822-3061. Circle 149.

(continued on page 152)



HIGH PERFORMANCE ISOLATOR ADAPTORS IN COMPACT PACKAGES

Cover the freq. spectrum from 2.6-18 GHz

- Standard models cover most current bands in use.
- Special designs adaptable to your needs.
- In production for Satcom, Point to Point, Radar etc.

FREQUENCY RANGE (GHz)	WAVEGUIDE SIZE-NUMBER WR	STANDARD FLANGE STYLES
3.7-4.2	229	CMR
4.4-5.0	187	UG
5.9-6.5	159, 137	CMR, UG
7.0-8.4	137, 112	CMR, UG
10.7-11.7	90, 75	CMR, UG
11.7-12.7	90, 75	CMR, UG
14.0-14.5	75, 62	UG

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Circle 151.

K_U-BAND WAVEGUIDE DIPLEXER

Part No. 601175 is a fixed tuned waveguide diplexer for K_U-band. Performance specifications include a transmitter frequency of 14.4-15.2 GHz, receive frequency of 13.6-14.0 GHz and SWR of 1.2:1 maximum. Unit has an insertion loss of 0.5 dB max., transmitter arm and 1.0 dB, max. receiver arm. Rejection-transmit to receive arm is 70 dB min. Waveguide size is WR62. Price: \$1150 each. Avail: 6-8 wks. ARO. **Coleman Microwave Co., Edinburg, VA. George Brinkley, (703) 984-8848.**

Circle 154.

DOUBLE BALANCED MIXER COVERS 0.5-18 GHz BAND

Model MD-166 is a double balanced mixer which spans the 0.5-18 GHz frequency band. Unit exhibits a 8 dB typ., 10 dB maximum conversion loss, an IF bandwidth to 5 GHz, 1 dB conversion loss flatness and isolation better than 20 dB on all ports. Other features include a 3-pin hermetic module for stripline and microstrip integration. SMA connectorized version also can serve as a module test fixture. Mixer can be screened to MIL-STD-883. Price: \$550 for module in 1-5 qty; \$625 for SMA model. Del: from stock. **Anzac Division, Adams-Russell Co., Inc., Burlington, MA. Mark Rosenzweig, (617) 273-3333.**

Circle 146.

SINGLE POLE, SINGLE THROW DIODE SWITCHES

Series of single pole, single throw diode switches cover the .1-18 GHz frequency range in bandwidths up to 9:1. Broadband isolations of 45 dB and 70 dB are available with and without drivers in switching speeds of 15 and 350 ns. Size: .69" x .75" x .5" without driver and 7/8" x 1.5" x .5" with driver. Price: From \$135. Del: From Stock. **Norsal Industries, Central Islip, NY. Norm Spector, (516) 234-1200.**

Circle 170.

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Circle 171.

MM-WAVE PHASE LOCKED FREQUENCY MULTIPLIER

Frequency multiplier Model 34.8X348-1 provides an output of 34.8 GHz at +20 dBm from an input of 100 MHz at +10 dBm. Loop automatically acquires and locks to the input signal. Lock range is +300 MHz and output is provided with an isolator. Options include source, lock indicator lights and metering. Price: \$8,500. Del: 120 days ARO. **Pacific Research and Development, Montrose, CA. John W. MacConnell, (213) 240-8510.**

Circle 172.

FREQUENCY MULTIPLIER TO 7200 MHz

A frequency multiplier which delivers 7200 MHz from a 450 MHz input is offered. Output level is 0 +1 dBm with an input of 23 +1 dBm. Output SWR is 1.5, max., noise in a 35 Hz BW is -77 dBc, min. and spurious and harmonic outputs are -50 dBc, min. Input and output impedances are 50 Ω. Size: 1.45" x .350" x .200". **KW Engineering, Inc., San Diego, CA. Art Brand, (714) 571-8444.**

Circle 145.

(continued on page 154)

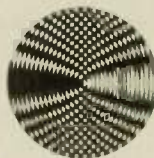
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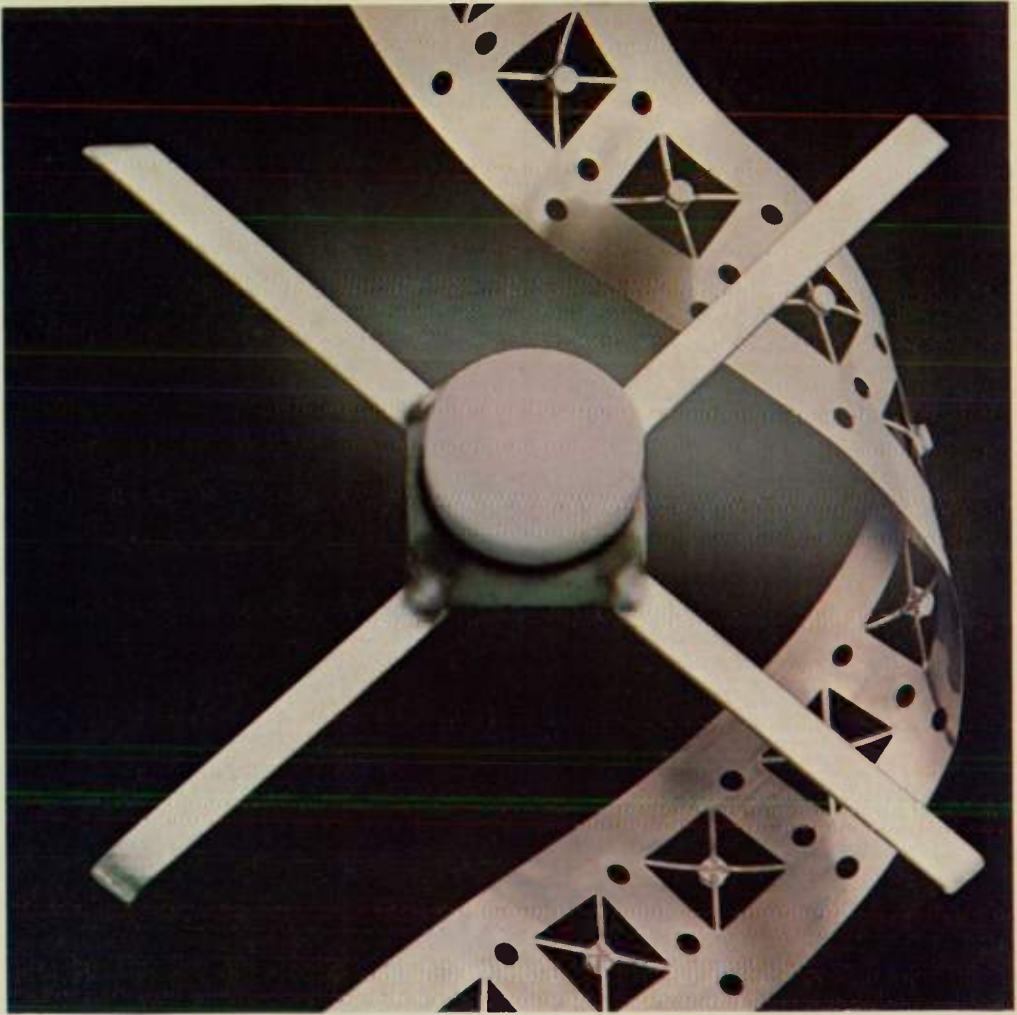
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The PA-5007 is a programmable attenuator designed for 10 bit Bdc input at 12 V and 15 ma per bit. The 10 bit code selects attenuation from 1 to 125 dB at frequencies from dc to 1.3 GHz. Per cell accuracy at 1000 MHz is 0.2 dB and .3 dB or 2% at 1.3 GHz. SWR is 1.3 at 1000 MHz; unit will handle 750 W, 3 s pulse and 0.5 W, average. Provided with Series D male connector for control inputs and SMA female RF connectors. Size: 1 1/32" x 2 13/32" x 5 1/8". Price: \$390, 1-9 qty. Del: 8 wks ARO. **Texscan Corp., Indianapolis, IN. Raleigh B. Stelle III, (317) 357-8781.** **Circle 174.**

HIGH POWER TERMINATIONS HANDLE UP TO 50 WATTS

Broadband, high power coaxial terminations are designed for applications from dc to 8 GHz. Model 1426 has a maximum average power input of 50 W and Model 1427 is rated at 25 W. Peak power rating of 5 kW is offered for both units. SWR is 1.20 to 4 GHz and 1.30 to 8 GHz. Connectors are semi-precision Type N male or female. Both terminations are constructed of a black aluminum body with an array of heat radiating fins. Price: 1426 - \$175; 1427 - \$100. Avail: Stock to 90 days. **Weinschel Engineering, Gaithersburg, MD. Julian Parker, (301) 948-3434.** **Circle 176.**

PIN DIODE ATTENUATOR

A 2-8 GHz, 64 dB minimum linearized pin attenuator, Model TG-1131, has SWR of 1.8 maximum, and insertion loss of 3.0 dB maximum. Control voltage is 0-7 V and sensitivity is 10 dB/V. Overall accuracy of the unit, including linearity, frequency flatness, and a -54 C to +85 C temperature range is ± 1.5 dB to 30 dB, ± 2.5 dB to 64 dB. Avail: 8 wks. **Triangle Microwave Inc., East Hanover, NJ. Martin Rabinowitz, (201) 884-1423.** **Circle 175.**

SMA QUICK-DISCONNECT RACK AND PANEL ADAPTER

An SMA quick-disconnect rack and panel plug adapter to SMA female is offered. The 705535-10 has a spring-loaded float mount and is intended primarily for blind mating rack and panel applications. The plug mates with a standard SMA jack. Connector meets all applicable portions of MIL-C-39012 and has a maximum SWR of 1.25 to 18 GHz. **Cablewave Systems, Inc., North Haven, CT. Steven Raucci, Jr., (203) 239-3311.** **Circle 148.**

120 K GaAs FET ANTENNA-MOUNTED LNA/DOWNCONVERTER

ACA-4220 is an integrated combination of a 3.7-4.2 GHz, 120 K GaAs FET preamplifier with a block downconverter to the 940-1440 MHz IF range in a case designed for antenna feedpoint mounting. LNA/downconverter features a minimum of 57 dB overall gain with ± 1.0 dB full-band gain flatness, and a +10 dBm typical output intercept point for intermodulation products. Assembly incorporates a dielectrically stabilized, 5140 MHz fundamental local oscillator. It features -45 dBm maximum LO radiation at both the input and output, and 60 dB image suppression (6.1-6.6 GHz). Input is a standard CPR-229 waveguide flange and it is powered from low voltage dc (15-28 Vdc, 250 mA) through the antenna-to-receiver feedline. Price: \$2,000 per unit. **Avantek, Inc., Santa Clara, CA. Don Smith, (408) 727-0700.** **Circle 147.**

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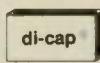
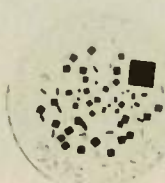
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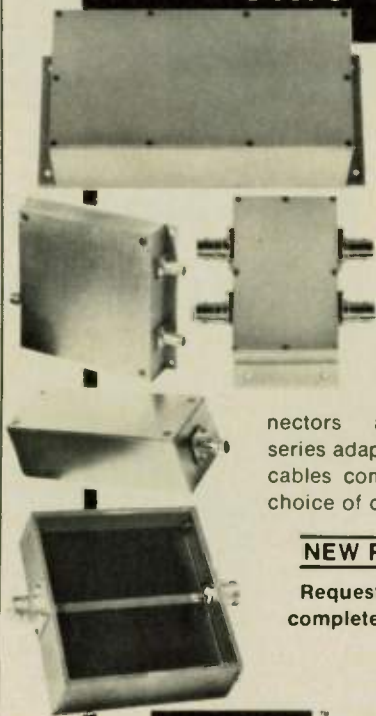
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Devices

SERIES OF Si SCHOTTKY DIODE QUADS

A series (ND487) of silicon Schottky diode quads are designed for high volume applications. Both low turn-on, N-type Schottky barrier quads and medium-barrier quads are available. Series is available in both interconnected ring and crossover configurations, and in a choice of MICRO-X (hermetically sealed stripline) or DISC-MOLD (plastic molded) packages. **California Eastern Laboratories, Inc., Santa Clara, CA.** Neil Corporon, (408) 988-3500. **Circle 143.**

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Antennas

ANTENNAS FOR MM-WAVE TECHNOLOGY

Equipment capable of producing accurate paraboloids, hyperboloids and ellipsoids for mm-wave technology yields low cost models available with several reflective surfaces. These include aluminum foil and copper, backed with epoxy/fiberglass and honeycomb. **Custom Radomes & Reflectors, Inc., Santa Clara, CA.** J. W. Downs, (408) 727-4777. **Circle 144.**

CIRCULARLY POLARIZED HORN WITH > 3:1 BANDWIDTH

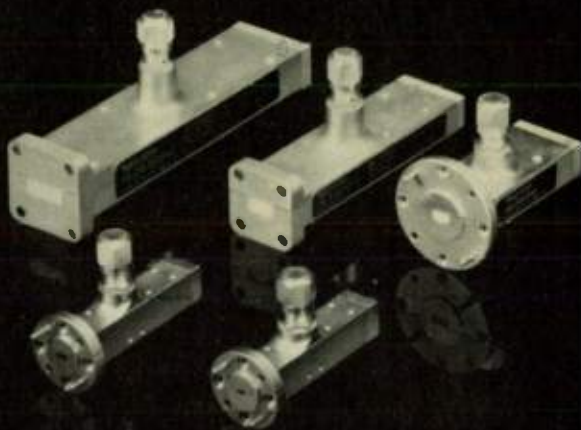
A circularly polarized horn with a 3.2:1 bandwidth capability is offered. Typical specifications for the antenna include SWR of less than 1.8, circular polarization (left or right hand), axial ratio less than 4 dB over 3 dB beamwidth. Lenses are available for providing a variety of beam widths. Coaxial or waveguide input is offered. **Adams-Russell Company, Inc., Amesbury, MA.** David Ryan, (617) 338-5210. **Circle 167.**

PORTABLE ANTENNA COVERS 1-12 GHz BAND

Model WJ-8593-24 is a 3 ft. diameter portable antenna that operates over the 1-12 GHz frequency range. This segmented-reflector antenna is designed to be easily disassembled and stored in a suitcase sized container. A dc-actuated coaxial switch at the rear of the crossed log-periodic feed allows the operator to select vertical or horizontal polarization. Minimum gain is 12 dB at 1 GHz and 32 dB at 12 GHz. Antennas has SWR less than 3.0:1 across the band and electrical performance is useful from 0.5 to 18 GHz. Reflector is made of fiberglass; the feed assembly is encapsulated in a foam-filled fiberglass container; captive hardware is used to connect the sections of the parabola and mounting provisions are provided on the back of the reflector. **Watkins-Johnson Co., San Jose, CA.** S. B. Witmer, (408) 262-1411. **Circle 168.**

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Letters to the Editor

NRL PHASED ARRAY WORKSHOP

I enjoyed your review of NRL phased array workshop in the February 1981 *Microwave Journal*. Has a Proceedings of the Workshop been issued and do you know if it has an NTIS or OOC number — either in the unclassified or unclassified version? Thank you for your assistance in this.

SAMUEL SENSIPER, Sc.D., P.E.
Culver City, CA

Editor's Note: Proceedings for the 1980 NRL Phased Array Technology Workshop should be ready in June, 1981. Security clearance is required for these classified proceedings and they are available at no cost. Contact: Mr. J. Paul Shelton, Code 5370, Naval Research Laboratory, Washington, DC 20375.

I enjoyed very much reading Dr. J. F. White's excellent summary of the NRL "Phased Array Technology Workshop." There are a few items that are worth commenting on.

First, the PAVE PAWS thinned array does use terminated dummy elements. What was stated is that we now know how to design thinned arrays without the use of potentially expensive dummy elements.^{1,2} This can be done with a small loss in gain and little effect on the antenna sidelobes. Future thinned arrays may be built without the dummy elements if the cost saving from leaving them out is greater than the cost accrued in making up for the loss of antenna gain.

Second, the COBRA DANE has 34,768 elements.

Third, the Grumman Aerospace Corp. space based array radar that Dr. James Frazer (DARPA) spoke of was a joint Raytheon/Grumman concept.

1. Beltran, F., "Elimination of the Dummy Elements in Thinned Phased Arrays," Final Report on Contract DASG60-78-C-0124 for BMDATC, Huntsville, AL, 23 August 1979, AD B043885L.
2. Beltran, F. and F. King, "Elimination of the Dummy Elements in Thinned Phased Arrays," to be presented at the 1981 APS Symposium, Los Angeles, CA, June 15-19, 1981.

ELI BROOKNER
Consulting Scientist
Raytheon Co.
Equipment Development Labs
Wayland, MA

ON "WERE YOU ZAPPED"

Your recent article in *Microwave Journal* presented very emphatically the common response to the question of possible biological effects of radiation. I looked briefly at the "Zapping of America" when it frightened one of my research associates. I am in general agreement with your assessment of the quality of the book. However, I was distressed that you did not mention the large body of work which explores the possibility of non-linear effects of microwaves at low intensities where thermal effects are excluded. Frohlich's recent review* suggests that biological systems can "store" the energy from microwaves until it "triggers" events for which the biological system is already prepared. If you can refute this large body of peer-reviewed work, I would be glad to see an article by you. Otherwise, I think you should have acknowledged that the issue isn't quite as simple as you made it seem.

* The Biological Effects of Microwaves and Related Questions. H. Frohlich, *Adv. Electronics and Electron Physics* 53, 85-152 (1980).

GARETH R. EATON, Professor
U. of Denver
University Park
Denver, CO

Editor's Note: Yes, there have been some non-thermal effects described, one of which was cited in, "Were You Zapped," the Frey effect by which some people can "hear" pulsed radar signals; however, to our knowledge none of these appears to be any more dangerous than, for example, exposure to artificial light, coffee drinking, etc.

Your editorial on zapping was timely and quite interesting. However, I was somewhat disappointed that it seemed to espouse the conservative doctrine which has been valiantly but unsuccessfully defended by John Osepechuk and others. Although much Russian and East European work in this area is suspect, there is now considerable evidence that there are significant effects which are neither ionizing nor thermal. In fact, COMAR is in the process of reducing the 10 mw/cm².

Reprints from the London ONR publication concern the recent URSI session on biological effects. Some other indications can be found in the journal, *Bioelectromagnetics*, which initiated publication in 1980. For example, in Volume 1, Number 1, Page 35, some people from the EPA Health Defects Research Lab report on an interesting non-linear power density window which affects calcium ion efflux from brain tissue at levels much below 10. Another paper in the same issue from the Naval Medical Research

Institute, on Page 89, discusses rat modified response scores for power densities as low as .25. The situation is confused by the fact that all these effects seem to be non-linear in power and there may be frequency resonances as well. Nonetheless, there is now a large accumulation of reputable data showing that there can be significant long-term effects at power densities below 1 mw/cm². On the other hand, I have no doubt that microwave ovens are completely safe and that most people are not exposed to any significant hazard.

ROBERT C. HANSEN
Consulting Engineer
Tarzana, CA

There is an invitation in your editorial to act as spokespeople in cocktail parties to disseminate sound information about potential microwave hazards among the general public. This I have been doing for years, not at cocktail parties, which I do not attend very frequently, but through my personal contacts, technical meetings, meetings with workers' organizations, meetings with the population involved with the installation of a new radar, contacts with the local administrations, local health offices, magistrates, military people, local newspapers and university lessons.

My contribution to the solution of the problem is this: with a little discernment and no additional cost it is possible to comply with even the most restrictive standards existing in the world, thus reducing the drama problem to an easily manageable level. In other words: through the problem of understanding the biological mechanisms of interaction of RF with living bodies certainly exists, the emotion on the subject of the risks of exposure can be reduced to practically nothing by showing that *levels above the most restrictive safety values are practically never found.*

The above statement is based on the results of several surveys which I was requested to perform with my collaborators on various potentially RF-polluted ambients. This usually occurred in connection with cases of public fear. In each case a written report was issued. The Introduction and Conclusions of each report were worded such that could be read by non-technical persons, while the body was in a form suited to an expert in the field. The surveys include:

- a) a TACAN transponder
- b) an ATC high-power radar
- c) a short-wave dielectric-heating machine for the furniture industry

-), e), f), g) others, as above
 n) a medium-wave induction heating machine
) two AM broadcast transmitters on the roof of an elementary school
) a high-power AM broadcast transmitter (100 kW carrier + 50 kW peak modulation)
 m) short-wave and microwave diathermy appliances in a room
 n) an anti-aircraft missile site with 5 radars.

PROF. L. MILLANTA
 IROE-National Research Council
 Firenze, Italy

Your article in the *Microwave Journal*, January 1981, is an excellent summary of the phony microwave radiation scare which has been promoted in recent years.

In answer to your question: "Was I zapped?" - No! Despite being exposed along with a few other technical people in 1944 and 1945 to power levels which were orders of magnitude higher than the milliwatt levels which have been feared in recent years.

For more than a year I worked with a group of MIT Radiation Laboratory which was responsible for upgrading transmission line components at S, X and K Bands to prevent high power breakdown. The procedure was to establish an arc, determine its location and redesign for higher power handling. We used high power equipment, 125 kW peak power at K Band, 250 kW peak at X Band and 1 megawatt at S Band, with a .001 duty cycle. The component to be tested was placed in a bell jar and terminated by a horn antenna so that we could see the arc when it occurred and determine its exact location. Generally, we set the power as noted above and decreased pressure until the arc occurred, while

ERRATA

Monolithic GaAs IC development work was incorrectly attributed to Marconi Research Laboratories in the Special Report, "The European Microwave Industry," *Microwave Journal*, January 1981. This work is being done at the General Electric Company Ltd. Hirst Research Centre.

Mon. M. Claude Chekroun is an employee of the Société d'Etude du Radant in Orsay, France and not a member of the staff of Radant Systems, Inc. in Stowe, Mass. as printed in the February, 1981 issue of *Microwave Journal*, page 45.

one member of the team looked into the horn and observed the position of the arc. The viewer moved away from the horn when his face became too hot to bear. The results of several hundred tests are presented in Radiation Laboratory Report No. 1071, dated November 1945, entitled "Summary of High Power Breakdown Tests on Microwave Components," by H. F. Clarke, G. L. Rogan, Richard M. Walker and Ina Mansur. I am not aware of any health problems caused by this work 35 years ago.

The waveguide horns used were of the order of 4 x 6 inches = 24 in² or

150 cm². Therefore, the peak power density was 830 watts/cm² at K Band 1670 watts/cm² at X Band and 6700 watts/cm² at S Band. The average power levels were .83, 1.67 and 6.70 watts/cm² at the horn face.

This information may be of interest to *Microwave Journal* readers.

RICHARD M. WALKER
 Founder Director
 M/A-COM, Inc. Companie
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 Burlington, MA

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Honeywell is looking hard for more people with that experience. Otherwise our person could be a BS/MS or Ph.D. in EE, physics or mathematics. Mild addiction to skiing or sailing weekends or an interest in tailgate-partying with colleagues at Minnesota Kicks or Minnesota Vikings games would be helpful but not required.

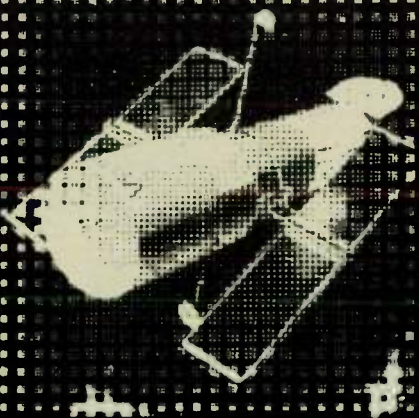
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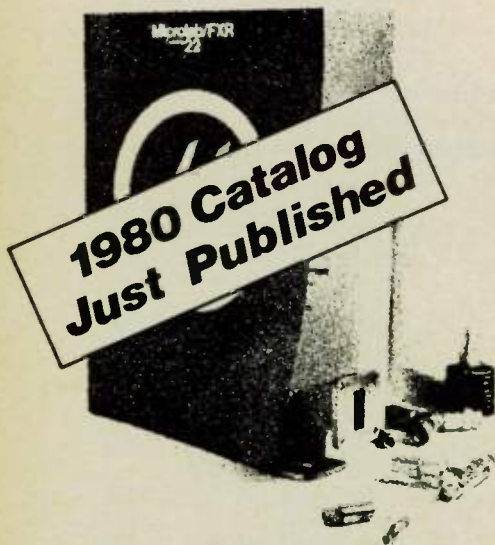
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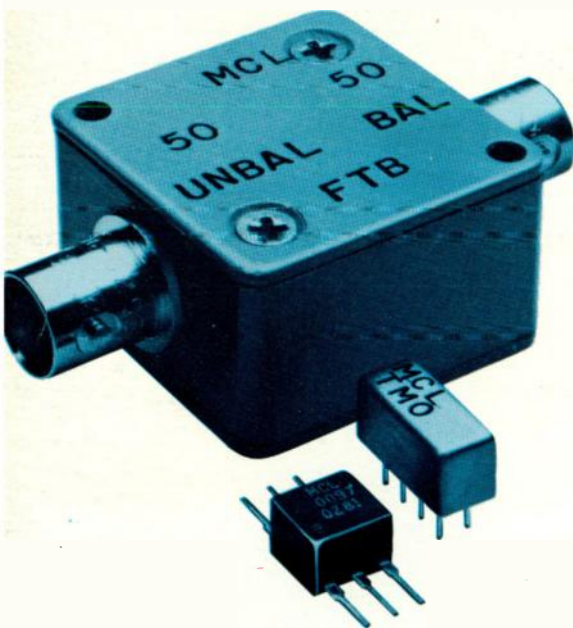
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Imped. Ratio	1	1	1.5	2.5	4	9	9	16	16
Freq. (MHz)	15-400	8-300	.1-300	.01-100	.02-200	15-200	2-90	3-120	7-85
T Model (10-49)	\$2.95	\$4.95	\$3.95	\$3.95	\$3.95	\$3.45	\$5.45	\$3.95	\$5.95
TMO model (10-49)	\$4.95		\$6.75	\$6.45	\$6.45	\$6.45		\$6.45	

CENTER-TAPPED DC ISOLATED PRIMARY & SECONDARY



	T1-1T	T2-1T	T2.5-6T	T3-1T	T4-1	T4-1H	T5-1T	T13-1T
Model No.	TMO1-1T	TMO2-1T	TMO2.5-6T	TMO3-1T	TMO4-1		TMO5-1T	TMO13-1T
Imped. Ratio	1	2	2.5	3	4	4	5	13
Freq. (MHz)	.05-200	.07-200	.01-100	.05-250	2-350	8-350	3-300	3-120
T Model (10-49)	\$3.95	\$4.25	\$4.25	\$3.95	\$2.95	\$4.95	\$4.25	\$4.25
TMO model (10-49)	\$6.45	\$6.75	\$6.75	\$6.45	\$4.95		\$6.75	\$6.75

UNBALANCED PRIMARY & SECONDARY



	T2-1	T3-1	T4-2	T8-1	T14-1
Model No.	TMO2-1	TMO3-1	TMO4-2	TMO8-1	TMO14-1
Imped. Ratio	2	3	4	8	14
Freq. (MHz)	.025-600	.5-900	2-600	15-250	2-150
T model (10-49)	\$3.45	\$4.25	\$3.45	\$3.45	\$4.25
TMO Model (10-49)	\$5.95	\$6.95	\$5.95	\$5.95	\$6.75

FT FTB



	FT1.5-1	FTB1-1	FTB1-6	FTB1-1.75
Model No.				
Imped. Ratio	1.5	1	1	1
Freq. (MHz)	1-400	2-500	.01-200	5-500
(1-4)	\$29.95	\$29.95	\$29.95	\$29.95

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