

73 Magazine for Radio Amateurs

- 34 Power Line DXCC (Distant Control Circuit)**
—the ac lines are already there... why not use 'em for remote control? **W9CGI**
- 42 Blueprint for Biofeedback Experimentation**
—fertile ground for pioneers **WD5BNL**
- 44 The History of Ham Radio**
—part X **W9CI**
- 48 Maximum Security for the 22S**
—remote control puts your Icom in the trunk **K8KW**
- 50 The Big Blinker**
—a visual signal for the deaf... **AD5X**
- 52 Morse Converter for DMMs**
—super gadget for blind hams **WA6AXE/3**
- 56 How to Home-Brew Your Own Crystal Filters**
—the series-string method. **Staff**
- 60 Four Bands on a Bamboo Pole**
—try a Chinese vertical slanter on 10 through 40. **W0VM**
- 64 The Triton IV Goes QRP**
—simply and efficiently. **W1FK**
- 68 Experimenter's Corner: The MM5369N**
—how many uses can you find for this one-chip crystal oscillator?.... **Patten**

- 72 Digital Readout Rotator Control**
—can the microcomputer connection be far behind? **K8TMK**
- 80 Build the KIM Keyer**
—with a 3-message memory **K0EI**
- 96 No More TRS-80 Cassette Woes**
—E-Z Loader does the trick **WA9PUL, K9POX**
- 100 The Incredible Shrinking Transceiver**
—build it on two tiny PC boards **K4DHC**
- 106 In Quest of Perfect Break-In**
—well... almost perfect **WB7CMZ**
- 116 The Amazing Audio Elixir**
—this limiting amp is a cure-all **N6WA**
- 118 No More Rotary Switches**
—gadget freaks will love these solid-state replacements. **WA2FPT**
- 122 Confessions of a Teenage Hfer**
—the world between channel 40 and 28 MHz. **Peter**
- 128 70-Watt Shoes for the IC-502**
—put some punch in that 6m signal **WA1PDY**
- 136 The Big Bopper**
—a 5-band phased vertical array **W0VDJ**



Never Say Die—4, Looking West—10, DX—12, Letters—14, RTTY Loop—20, New Products—22, Microcomputer Interfacing—26, Awards—28, Contests—30, Ham Help—147, 151, 155, Review—154, OSCAR Orbits—154, Corrections—155, Social Events—159, Dealer Directory—159, Propagation—193

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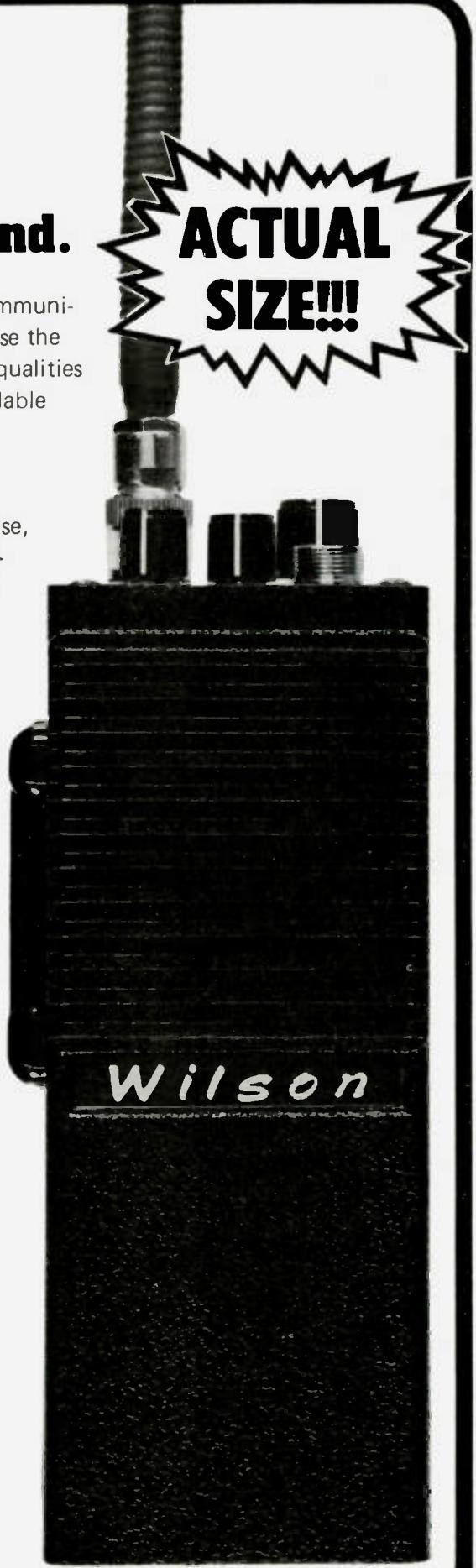
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W2NSD/1 NEVER SAY DIE

editorial by Wayne Green

10-GHZ RECORD EXTENDED

Not satisfied with holding the world's record for states worked on the amateur 10-GHz band (five), the record holders have been working to extend it to six . . . perhaps seven. The sixth was worked, I'm glad to report.

It all started out last fall when Chuck Martin WA1KPS got interested in the Microwave Engineering Gunnplexers and got a pair of them perking. The first contact from New Hampshire to eastern Massachusetts was relatively easy. Oh, it took us about fifteen minutes to find each other while tuning one unit and sweeping the general direction of the hill a bit over 50 miles away. Eventually, we lucked onto both the right frequency and bearing and the contact was solid.

Later, Chuck made an expedition to a mountain in southern Maine, also well over 50 miles away. Again it took a lot of tuning and turning of the two transceivers before the contact was made. Both have to be on the right channel and pointed very close to the right direction. No dishes were involved at this time—just the Gunnplexers, with their little waveguide antennas.

Chuck then drove to Mt. Ascutney in Vermont, again over 50 miles from my mountain in southern New Hampshire, Pack Monadnock. It is no accident that I have a drive-up mountain within four miles of the 73 offices . . . that's what attracted me to this area when I moved here, 17 years ago. I've been a VHF and UHF nut right from the earliest days of my hamming, starting out on 2½ meters, for those of you who remember that band. I ran a pair of 76s with their bases slotted for "efficiency" and was the terror of Troy (NY). That was before WWII . . . forty years ago!

Next we needed to work New Hampshire, so Chuck drove to the top of Mt. Washington and we made it . . . but that one was

tough. We were both freezing in the cold winds, clouds were swirling around Washington, and the transceivers were not working well at all. In the light of later improvements, they had never been working more than marginally! But we did make it for an exchange of callsigns and signal reports . . . whew! That was 106 miles, with faulty transceivers and no antennas to boost the signal.

It took two tries to get Rhode Island, and we had several frustrating failures with Connecticut. Just as Chuck had the transceivers working better and a two-foot dish ready to use, snow finally blanketed my mountain and that was that for the winter. When spring finally arrived, we tried to make it to Connecticut again, with Chuck even climbing trees to try and get the few extra feet of height needed. No go. Then we tried to make it through from the New York border . . . again a failure. There was no strain when we tested the system out between the Pack and Mt. Greylock in western Massachusetts, but the path to New York was just not there.

Bigger problems call for sterner measures. The Pack is about 2,500 feet high and is usually excellent for VHF contests. It has the benefit of a toll road to the top . . . the very top . . . plus a fire lookout tower for a few added feet. About 12 miles away is Mt. Monadnock, a 3,500-foot lump whose top half is all rock. This mountain has a slight problem in that it is a 2½-mile hike to the top and the going is not easy. I'm almost 60 years old, my primary exercise is sitting at a typewriter all day, and I have no business climbing a 3,500-foot rock mountain. A few days of skiing during the winter helps a bit, but does in no way prepare one for hard mountain climbing.

Put into perspective, world records call for that extra effort, so I agreed to make the climb.

That was not one of my better decisions. Chuck rounded up his team of Steve K1KEC and Eric WA1HON and headed for the best possible location in Connecticut . . . near Somers. I rounded up our summer tech, Tim Daniel N8RK, and Sherry and we headed for Mt. Monadnock.

The first half was easy . . . a dirt road. The second half was a pain. There was no scaling of rock faces—it was just a grueling climb up rocks for over a mile. It seemed to go on forever. We'd see what looked like the top just above us maybe 500 feet, but when we got there we would see another peak beyond that. Sherry, who is used to mountain climbing being like walking up Stone Mountain in Atlanta, said that this was between ten and a hundred times as difficult. Tim? At 17 years old, he went up like a mountain goat.

When I finally arrived, panting and a bit worse for wear at the top, I was greeted by a couple hams and a three-element six-meter beam! They'd packed 40 pounds apiece up there for some DXing on six meters, only to find a maximum of six contacts to be made. It was Sunday afternoon, with very little activity on six meters. There were a couple dozen hikers, all looking disgustingly fresh and rested.

Tim and I set up the 10-GHz rig and aimed it in the direction of Connecticut, coordinating the effort on 220-MHz simplex and, on occasion, on two-meter simplex. This was the easiest of them all . . . contact was made almost immediately. Chuck's group was using a two-foot dish on top of a fire tower and we had the plain Gunnplexer, but mounted on a tripod this time for stability. The signal strength pin went right over full scale and then off scale! The voice was just like the other team was right with us. We did get some interference from the six-meter team breaking through on our



TR-7600



TR-7625



RM-76

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Featuring 25 watts RF output (switchable to 5 watts low power), the TR-7625 is a high-performance 2-meter FM transceiver with memory, and is designed to permit multi-channel (800-channel) operation. Compact and perfect for mobile or ham shack use. When used with optional RM-76 Microprocessor Control Unit, the TR-7625 offers a whole new dimension in channel memory and scanning capability.

TR-7600

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RM-76

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TS-700SP

SPECIFICATIONS	Model: TR-7600/TR-7625*	Model TS-700SP	Model TR-830
Frequency Range:	144.00 to 147.995 MHz	144.0 to 144.0 MHz	TX: 445.0 to 440.0 MHz RX: 442.0 to 447.0 MHz
Modulation:	FM	SSB (USB, LSB), CW, AM, FM	FM
Dimensions:	131mm (5-1/16") wide 61mm (2-3/8") high 290mm (5-1/16") deep	278mm (10-7/8") wide 124mm (4-7/8") high 320mm (12-5/8") deep	180mm (7-1/16") wide 60mm (2-3/8") high 240mm (9-7/16") deep
Weight:	1.75kg (3.85 lbs) Approx	1.1kg (2.42 lbs)	2.3kg (5.1 lbs)*
RF Output Power:	High: 14 (or 25) watts (min.) Low: 1 (or 5) watt approx (adjustable to 10 watts)	SSB, FM, CW—10 watts AM—3 watts FM (Low)—Approx. 1 watt	High: 10 watts Low: 1 watt Approx
Modulation:	Variable reactance direct shift	SSB: Balanced modulation FM: Variable reactance frequency shift AM: Low power modulation	Variable reactance phase shift
Microphone:	Dynamic microphone with PTT switch, 500Ω	Low-impedance microphone (500Ω) with PTT switch	Low-impedance microphone (500Ω) with PTT switch
Sensitivity:	Less than 0.4µV for 20 dB quieting	Less than 0.4 µV for 20 dB quieting SSB & CW: 0.25 µV for 10 dB (S+N)/N AM: 1.0 µV for -10 dB (S+N)/N	1 µV to 30 dB (S+N)/N 0.5 µV for 20 dB noise quieting
Squelch Sensitivity:	Less than 0.25 µV	0.25 µV	0.3 µV
Selectivity:	More than 76 dB at 30 kHz or adjacent channel	SSB, CW & AM: ± 4 kHz/-6 dB 4.8 kHz/-6 dB FM: 12 kHz/-6 dB 24 kHz/-60 dB	20 kHz/-6 dB 40 kHz/-70 dB
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Peterborough NH 03458
Phone: 603-924-3873, 924-3874

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link. RFI, even on top of Monadnock!

The climb down was not much easier than going up. It didn't take as much grunting and straining up through the rough places, but it took a lot of care and knee-shaking jumps to rocks. I was lame for several days afterward with knee-cap shock. The thought that Chuck and his group might want to do this all over again for another try for the New York-New Hampshire path went through my mind with each and every rock that I climbed down... all one damned mile and a quarter of them.

Fortunately, the day was reasonably cool... perhaps around 80°... and beautifully sunny. Visibility was only about 30 miles in the haze, but then it was a natural haze caused by the trees and not smog. There is never anything but clean air in New Hampshire.

The team down in Connecticut didn't have to cope with climbing a rock mountain, but they did have their problems. Being considerably south of us, they had much hotter weather and suffered greatly from sunstroke and heat prostration. Steve became completely prostrated in the back of their van when the time came to disassemble their setup and carry it down the tower. Oh, the agonies we have to go through to become world champions.

BITING THE LIFERS

A recent letter from Chapman (ex-New England director) to ARRL life members on ARRL Foundation letterhead asked for donations to set up a \$100,000 slush fund for HQ to spend at Geneva this year.

For some reason, nothing was mentioned of the \$100,000 fund which has already been set up for just this type of use. I'm reminded of the League saving money for years to buy a new headquarters building... money taken from the members as dues... and then, when they finally had enough money, running a building fund and collecting the same money a second time. And to cap that, they sold their old building for almost the cost of the new building. That's mainly how they built up that \$2 million-plus they have in securities and banks.

It seems to me that this begging for more money calls for some whistle-blowing. They seem to be trying to make up for the recent bad management of the League by asking for donations in the name of WARC.

The \$100,000 is still carried on the balance sheet of the League in their latest annual report... it is for "the defense of amateur frequencies." That would seem to cover WARC like

a glove. The fund was set up years ago and has been used for undisclosed purposes and then refunded by the League at the end of each year. It would seem that members of the League are entitled to a detailed accounting of the money spent from this fund down through the years. Any publicly held corporation would have to account to the shareholders for such a mysterious fund... particularly when there is a serious question about it being used for the personal benefit of League officials. IRS, please note.

I think I join many members in hoping that the League will be more prudent at Geneva this year. At the last WARC, they had a huge suite in one of the most expensive hotels in town and entertained lavishly. Throwing money around like that is the Ugly American way and we need friends, not resentful enemies.

To sum it up... the ARRL Foundation is in disgrace over the resignation of most of its directors... a fact which HQ has refused to discuss or even let members know about. They already have \$100,000 in cash set aside for just such an event as WARC, so the begging for \$100,000 more stinks. If you have money to pour down rat holes, I'm sure you can find better charities than this one.

ARRL PROBLEMS

I see that the ARRL's Central Division convention was a monumental flop... eight exhibitors and about 400 in attendance. The magic seems to have gone out of ARRL conventions, and the active hams are staying away in record numbers. Compare this to the Atlanta Hamfest, where, in spite of the severe gasoline shortages, the attendance was only down a bit from last year and the exhibit hall was packed solid with exhibits and people. Compare it with the St. Louis hamfest, where a first-time show filled the exhibit hall and packed the hall with attendees.

What were the differences? One was a matter of promotion, where Atlanta and St. Louis got extensive promotion in both 73 and *MICROCOMPUTING* magazines... ads and editorial. Another difference was the speaker program, where one had no one of a controversial nature... not even of any significance... and the others had speakers who packed 'em in. I suspect that the day is past when you can pull in a good crowd at a hamfest just by listing it in QST.

OUTLAW RADIOS

A recent attempt to amend a New Jersey law which prohibits the use of radios in cars

capable of receiving police channels so that it would exempt amateurs was vetoed by Governor Byrne. Someone should make it a point to acquaint the governor with the facts of radio life and the ridiculousness of the whole charade.

Maybe a few years ago something like this could be legislated, but with the modern scanners there is no way to keep anyone who wants to from receiving police channels. Like guns, the laws only have an effect on the honest person. Criminals carry guns and are quite capable of buying and using radios which will receive police channels... and how can a policeman know? With just about every other car sporting some sort of special antenna, a car can be set up for CB, for hamming, for any of a hundred special services... or to illegally receive police channels.

If the police are looking for just one more rap to lay on a criminal once he has been caught, the ramifications for hams and other honest users of radios are much too great a penalty to pay. Virtually every mobile ham rig for VHF is quite capable of being tuned to a police channel. The scanners, which are being sold by the tens of thousands, can get just about every police channel there is, high- or lowband... and they are 12-volt jobs, just fine for mobile use.

Will someone please bring Governor Byrne up out of the 1950s? And perhaps someone should alert the New Jersey legislature that making it illegal to have radios capable of receiving police channels in cars is not going to have any effect on criminals... it will only harass non-criminals.

SPREADING THE WORD

One of the jobs that I feel a national amateur radio organization should do is lobby for us on a national level. This means organizing campaigns which will get amateur radio favorably into the media... articles in national magazines... demonstrations which are covered by television... promotion of ham activities during emergencies, etc. In case there is any doubt, we are not getting such a service from any of our national amateur radio organizations.

Individual clubs, in a few instances, have been doing a fine job of getting publicity, but unless you are terribly isolated from the general public, you must be aware that most people have no idea of the difference between CB and ham radio. We do need PR and we need it badly.

Continued on page 148


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Looking West

Bill Pasternak WA6ITF
24854-C Newhall Ave.
Newhall CA 91321

The last time I went on a Field Day outing was in the early sixties. I was a member of an organization known as the Flatbush Radio Club of Brooklyn, New York, and we took to the hinterlands of New York City. The place was Prospect Park, which sits in the center of Brooklyn. OK... so it wasn't really the hinterlands. It was fun, but through the years the memories of that outing grew faint.

Since moving to California, the closest I have come to taking part in a Field Day outing was the time I went from club to club showing the PARC Field Day film, *Field Day 1972*. In the seven and one-half years I have been here, something has always come up to make me miss Field Day. One year I was out of town, another time I had to work... you know the story. Well, this year it all clicked. The San Fernando Valley Amateur Radio Club was holding its Field Day operation on a hill not twenty minutes from my house. I had film in my Polaroid in addition to 600 feet of Super 8 film. Most important, I had the time and the opportunity to attend. The lure was overwhelming...

If you have never experienced a real Field Day operation, you've missed something important. This is true gut-level amateur radio operation at its

best. It's you the amateur versus the bands, the elements, and the odds. To me, the whole idea of Field Day is not whether I win or lose, but rather the camaraderie that is shown among amateurs during such an event. The San Fernando Valley Amateur Radio Club is an excellent example of such fellowship.

Let me tell you a bit about the SFVARC. It is a true amateur radio club with members whose interests go in many diverse directions. Unlike most clubs of this era, the SFVARC does not own or operate a repeater, nor does it ever plan to. In fact, a few years back I offered them such a device and it was rejected. It is not that SFVARC members don't like repeaters—far from it. Most of the members can be found operating the myriad of two-meter and 220-MHz systems that abound in this locality. However, to the SFVARC, a repeater is nothing but a way of extending the communications range; with over 300 operational 2-meter repeaters in southern California, there is no need for yet another. So, while most members do operate VHF or 2-meter FM, they do so using the existing relay facilities.

Rather than worry about running a repeater, the SFVARC looks in other directions. It does such things as provide communications for March of Dimes walk-a-thons and work with

"Santa" at Christmas to help bring joy to children in hospitals. The best way to describe the SFVARC is to say that they epitomize the concept of an amateur radio club in the truest and broadest sense. Perhaps that is why Field Day with SFVARC was so much fun. There were about a dozen dedicated people from the club spending almost thirty-six hours on a remote California hilltop enjoying each other's company and their amateur radio hobby. They were fulfilling a dream that could only be fulfilled once a year. Being together enjoying the experience was the point of their hilltop excursion. I could only spend Sunday with them, but it was well worth it. My memories of the past were rekindled and somehow I felt more like a "real ham."

A NEW MOVIE

Hollywood, which has brought you such epics as *The Poseidon Adventure*, *Two For the Seesaw*, *Charlie*, and other great moments in cinema, is now proud to announce another achievement. *The World of Amateur Radio* was produced by a man I consider to be a true cinema artist, Dave Bell W6AQ. I am writing this only an hour after leaving Dave's Hollywood studio. We just screened the first answer print of this new film and the consensus is that it is a superior work. With Dave Bell as the film's producer, the results were not surprising. In my opinion, he is a genius in his chosen profession. He knows his business, knows how to get a job done properly the first time out, and has the ability to lead others in the proper direction. Being able to work with someone like Dave has been a

real honor and a truly rewarding experience.

The World of Amateur Radio is fast-paced, relevant, and covers most aspects of our amateur radio world. NBC-TV News correspondent Roy Neal K6DUE is the film's host and chief narrator. There are contributions from Senator Barry Goldwater K7UGA, Arthur Godfrey K4LIB, and Stu Gilliam WD6FBU, and special appearances by Dick Van Dyke and His Majesty King Hussein of Jordan JY1. The film was produced by Dave Bell Associates for the American Radio Relay League and it will soon be available from the ARRL. I am happy to say that this is one time when the League is truly serving the amateur community; it should be applauded for its fine work on our behalf. Through the cinematic expertise of Dave Bell, the ARRL is providing us with the type of visual vehicle we need in order to explain what amateur radio is all about. As this is written, the film print is en route to Geneva, Switzerland, to be presented to those who will be deciding our future—the delegates to the World Administrative Radio Conference.

The World of Amateur Radio is our world, and now, thanks to Dave Bell and the ARRL, we will be able to share it with virtually everyone. All who worked on the project sincerely hope that you will enjoy the film. I have one request: At present, only one print of the film exists, and that print is in Geneva. It will be a few weeks before distribution prints are available. Please do not contact Dave Bell Associates or the ARRL requesting the film. Instead, keep watching for ARRL official bulletins, *HR Report*, the Westlink Amateur Radio News, and this column. All these sources will let you know the moment the film is ready for release. While we wait, readers of this column may want to start recruiting non-amateur groups for introductory talks about amateur radio, using the film as the focal point of the meetings. While we amateurs will enjoy viewing it, it's far more important that the film be seen by the non-amateur general public. It is time for the rest of the world to discover us and our service capabilities. Therefore, it is extremely important that the film reach as wide an audience as possible. Public schools, public television, civic groups, and church groups can all be approached. This is where you come in. The film will do no good sitting on a shelf. To be effective, it must be seen. Any takers?

Ever wonder how a film is produced? Here's the secret... teamwork! (Photo by Bill Orenstein KH6IAF.)



Continued on page 153

New OMNI/SERIES B Filters The Crowd

The new OMNI/SERIES B makes today's bands seem less crowded. By offering a new i-f selection that provides up to 16 poles of filtering for superior selectivity. And a new Notch Filter to remove QRM. No other amateur transceiver we know of out-performs it.

NEW I-F RESPONSE SELECTION. OMNI comes equipped with an excellent 8-pole 2.4 kHz crystal ladder i-f filter which is highly satisfactory in normal conditions. But when the going gets rough, the new OMNI/SERIES B, with optional filters installed, provides two additional special purpose i-f responses.

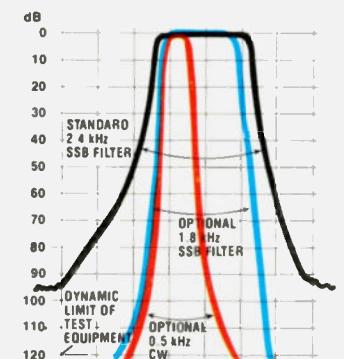
The 1.8 kHz crystal ladder filter transforms an unreadable SSB signal in heavy QRM into one that gets the message through. The 0.5 kHz 8-pole filter provides extremely steep and deep skirts to the CW passband window which effectively blocks out even the very strong adjacent signals.

Both of these filters can be front-panel switched in series with the standard filter to provide up to 16 poles of filtering for near-ultimate selectivity. In addition, the standard CW active audio filters have three bandwidths (450, 300, and 150 Hz) to give even further attenuation to adjacent signals. In effect, OMNI/SERIES B has six selectivity curves—three for SSB and three for CW. That's true state-of-the-art selectivity.

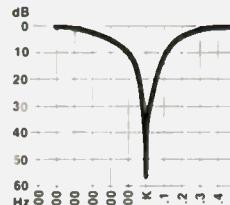
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NOTCH FILTER PERFORMANCE
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Model 545 Series B OMNI-A \$949
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LETTERS

INTERESTING METHODS

The article by Jerrold Swank W8HXR, "Compact Beams for 20 or 15," in the July, '79, issue provides some interesting methods of compacting HF beams.

Swank's method of feeding, however, may cause considerable exasperation to anyone who attempts to duplicate his results, or who picks up the tidbit about using a quarter wavelength of RG-59/U as a quarter-wave transformer providing a 2:1 step-up ratio and later attempts to apply the principle to a similar feed problem.

Swank states, "I used a quarter wavelength of RG-59/U as a matching section from the driven element to the RG-8/U coax lead-in. I guessed that the beam impedance would be about 20 Ohms, and this would raise it to 40 Ohms."

His first guess was probably close, but his guess about what the quarter-wave RG-59/U matching transformer would do missed by a country mile. The 20-Ohm feedpoint agrees closely with published data in several books by Orr and the ARRL Antenna Book, where driven element-to-director spacing approximates 0.11 wavelength spacing. Swank's 20-meter beam with an 8-ft. boom would provide this condition.

However, a quarter-wave matching transformer of RG-59/U (73 Ohms), will only provide a 2:1 step-up ratio where $Z_S = 106.07$ Ohms and $Z_R = 53.035$ Ohms. These conditions are not met, since he estimates Z_R to be 20 Ohms.

Given that his feedpoint resistance does, in fact, conform to textbook data, what his quarter-wave matching transformer of 73-Ohm coax will do is transform the 20-Ohm Z_R up to 266.45 Ohms, presenting then a 5.124:1 mismatch to the 52-Ohm RG-8/U feeder.

This can be proven mathematically by the standard quarter-wave transformer equations: $Z_S = Z_0^2/Z_R$ and $Z_0 = \sqrt{Z_R Z_S}$, where Z_0 = characteristic Z of line, Z_R = resistive termination, and Z_S = input Z of the line.

From the equations, we can also calculate the correct Z_0 to provide the desired match between 20 Ohms (Z_R) and 52

Ohms (Z_S), the nominal Z of the RG-8/U feeder. For these impedances, the Z_0 should be 32.2 Ohms, which is not a standard coaxial line. It can, however, be accomplished by using parallel quarter-wave sections of RG-59/U or RG-59B/U, connecting both braids and both center conductors in parallel. This will provide an equivalent line with Z_0 of 36.5 Ohms (75 + 2), a value close to the ideal. To calculate the mismatch into the RG-8/U using such a transformer, it is necessary to refer again to the equation: $Z_S = Z_0^2/Z_R$; $Z_S = 68.6$ Ohms.

Impedance mismatch between the RG-8/U (52 Ohms) and the quarter-wave transformer Z_S (68.6 Ohms), being a function of the ratio of these two values, then becomes 1.28:1, obviously a very acceptable value and a substantial improvement over the mismatched condition presented by Swank's quarter-wave of RG-59/U.

This then raises a question about how Swank obtained an SWR measurement of 1.5:1 at the lowest point. This would make it appear as though the quarter-wave of RG-59/U was transforming a Z_R of 68.32 Ohms to a Z_S of 78 Ohms, since RG-8/U terminated in a 78-Ohm load would operate with the 1.5:1 SWR he indicates. With a parasitic element spaced only 0.11 wavelength from the driven element, a feedpoint resistance of 68.32 Ohms would be much higher than normal, unless the elements were severely detuned. While this is possible, it is much more likely that his indicated 1.5:1 SWR is simply incorrect. This can be verified by adding another length of RG-8/U between the feeder and SWR bridge. The chances are that the SWR will then change, indicating that neither the 1.5:1 value nor the second value obtained present a true picture of what the mismatch actually is.

My purpose in writing this is twofold—first to correct the error in the article, since Swank's assumptions about the quarter-wave transformer were totally incorrect and may lead someone astray, and second, to demonstrate that his original guess of a 20-Ohm feedpoint resistance may also have been in error.

Before applying any method of improving the impedance

match, it would seem logical to actually measure the feedpoint resistance. Two of the regular advertisers in 73 offer economical equipment for this purpose: MFJ's MFJ-202 RF Noise Bridge and Palomar Engineers' RX Noise Bridge. Both will indicate feedpoint resistance in addition to the resonant frequency, eliminating the need to "guess" what these critical characteristics might be. To hang a matching system on an unknown feedpoint resistance is kind of like scooping a handful of pennies into a bag, then saying that by removing 20, only 52 will remain.

Robert G. Wheaton W5XW
San Antonio TX

The chances are that my feedline was not a real half-wave multiple. Anyway, the simplest solution would be to use a pair of 59/U coax cables in parallel, which would be 37.5 Ohms. The length required is only about 10' 10", and this would give a match of 1.4:1.

The alternative would be to use no matching section, which would give a match of 2.6:1, and, if necessary, trim the feedline for a match or use a small L network.

I wrote the article mainly to give hams ideas on how to make beams with the materials at hand. I apologize for the error.

Jerrold Swank W8HXR
Washington Court House OH

MINUTES

A little research of the Minutes of the ARRL Board meetings the past six years has turned up a few interesting facts and raised a few questions, all of which might be the subject of a future editorial: expense accounts of the Directors and the President of the ARRL.

Naturally, being in the Hudson Division, I took a look at the expense account of this Division Director. It looks like this:

	Authorized	Increase	Total Received
1979	\$4,000.	\$(note)	\$4,000. + ?
1978	3,500.	300.01	3,800.01
1977	3,000.	204.30	3,204.30
1976	3,000.	none	3,000.00
1975	3,000.	36.23	3,036.23
1974	2,500.	none	2,500.00

Note: The increase in authorization to a Director comes at the Board meeting the following year.

Now, I work for the State of New York, traveling extensively throughout the state (not just the Hudson Division), with an occasional extra trip to Harrisburg PA or to Washington DC. My expenses include lodging, subsistence, plane fare, airport limos, taxis, and car rentals. My expenses for any year haven't come anywhere near the Hudson Division Director's ex-

penses authorized for 1974, the minimum of those checked!

Another very interesting fact turns up when looking for the expense accounts authorized to the President. The 1979 Board (1st) meeting authorized Mr. Harry Dannals an amount not to exceed \$10,000 "for 1979." Looking at the Minutes again, I can't find any authorization for Mr. Dannals for the years 1978 and 1977. (Isn't that strange?) He was authorized \$7,500, "for 1976," though. Apparently nothing was authorized for 1975 and 1974. Question: If, indeed, as I suspect, the President did receive expense authorizations for the years 1978, 1977, 1975, and 1974, where were they hidden???

The above "gravy trains" are being financed by the membership through an increase in dues in 1977 to \$12 a year, and again in 1979 to \$18 a year. It is no wonder that they are finding it difficult to increase the membership in the ARRL at these rates.

A most significant item (Minute #74) in the Minutes of the January, 1979, Board meeting, makes me very suspicious of financial finagling in Newington. Mr. Holladay made this motion that the General Manager (Baldwin) be directed to publish the League's audited financial statements in QST. This motion was amended by Mr. Price (at whose instigation?) to have the editor of QST directed to remind members through "League Lines" of the availability upon request of the League's audited financial statements. The amended motion was adopted, over the opposition of Mr. Holladay.

It is no wonder, with all the "lulus" available, that the Directors and the President are not desirous of rocking the boat by asking for an investigation of blatant dictator management of the ARRL in Newington.

Byron H. Kretzman W2JTP
Huntington NY

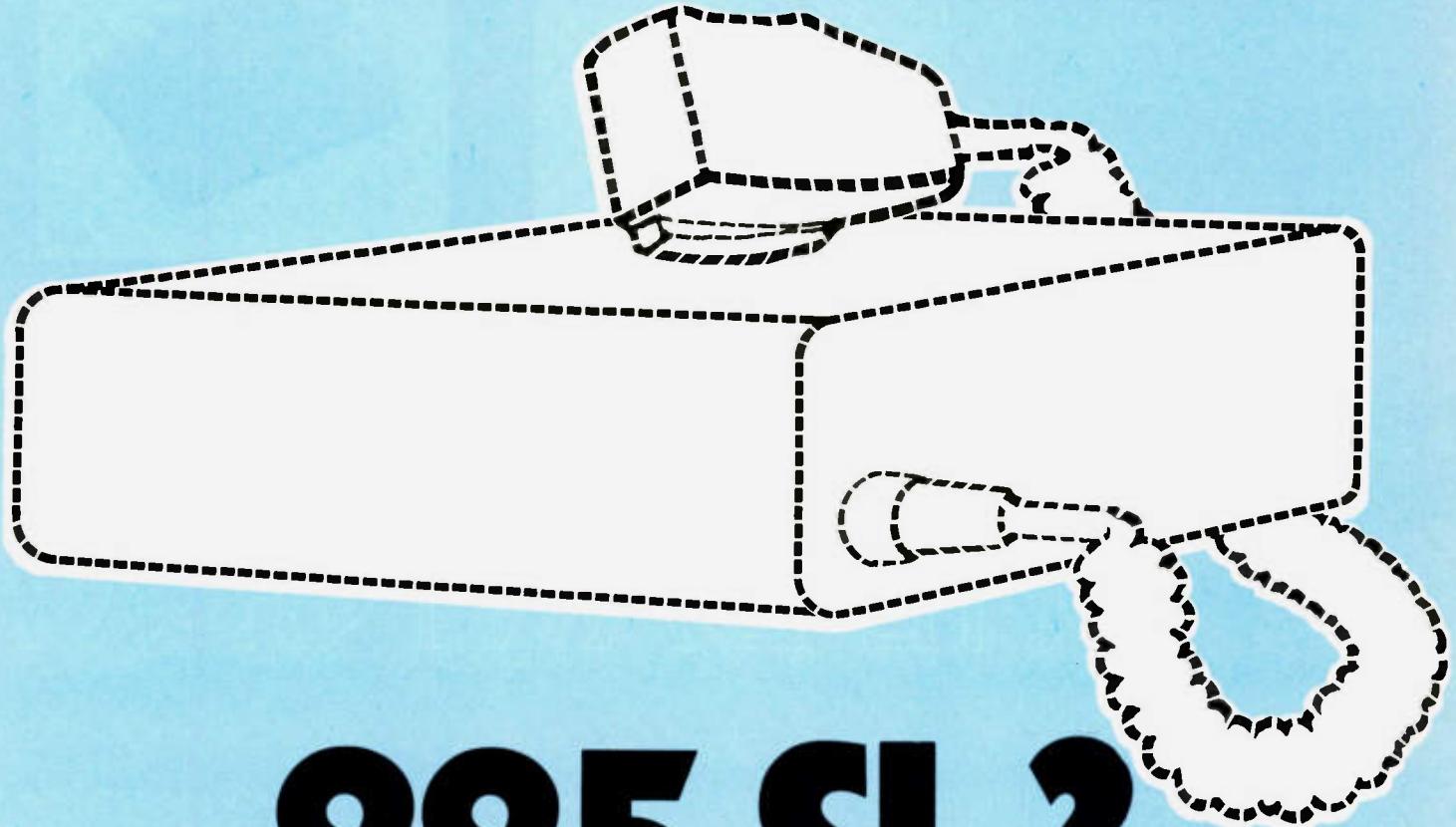
SACRIFICE

With mixed feelings, I wish to inform you that, in a drive to cut down my subscription list to something commensurate with my time to read, I am renewing your magazine and sacrificing the comparably laudable *Scientific American* and *Mother Earth News*. Yours has more circuits.

I have successfully resisted becoming a ham for some 35 years, and now that I am partially deaf (the National Semiconductor LM389 makes a dandy hearing-aid amplifier chip, though), it's unlikely that I'll ever succumb. I'm a longtime

Continued on page 156

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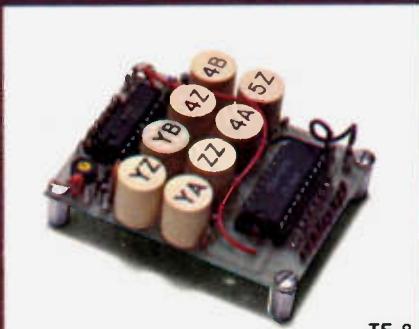
THE DAWNING

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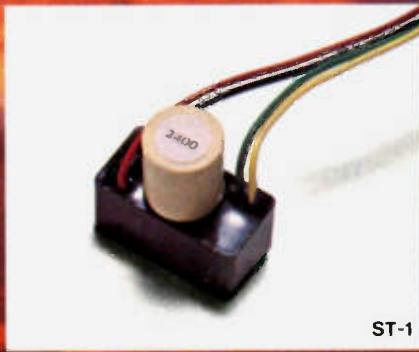
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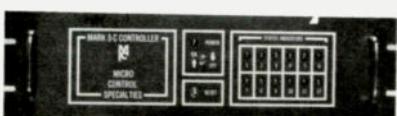
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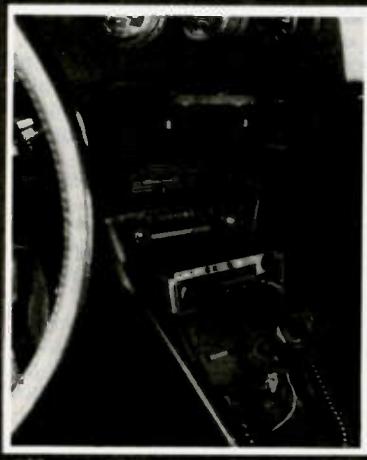
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RTTY Loop

Marc I. Leavey, M.D. WA3AJR
4006 Winfield Road
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Occasionally, a letter comes in which is so far afield from an idea I was trying to get across that I feel I must have made too many assumptions in the presentation. Such is the case this month.

Pete Buyaki K5GV is having trouble understanding my presentation of hexadecimal arithmetic. Pete writes, "[...] hit a snag when I came to \$5F, etc. This seems odd since these numbers are in the middle of the ASCII table, because it's in the middle of the alphabet table... Looking up in the ASCII tables:

\$ 0100100	
S 0110101	
F 1000110	
	Subtracting this group from the group above
\$ 0100100	space?
2 0110010	ETX (end text)
0 0110000	SYN (synchronous idle)
the difference	0000000
	0000011
	0110000

"I'm sorry, but I don't understand the relation as to how the Baudot will remain in the capital letter format when ASCII is putting out lowercase letters."

Okay, Pete, there are two items of note here. First off, the convention representing one eight-bit byte in hexadecimal is "dollar sign—left nibble—right nibble." In other words, 01001110 in binary corresponds

to 4E in hexadecimal. So, when I say \$4E, what I mean is 01001110. Now, in ASCII, that \$4E corresponds to "N". Lowercase "n" is \$6E, which is \$20 more than \$4E. That is, \$4E + \$20 = \$6E. Another way to look at that is by converting the hexadecimal to decimal. \$4E is 78 decimal, and \$20 is 32. Thus, 78 + 32 = 110, and 110 decimal is \$6E hexadecimal.

If everything is clear so far, I'll throw in one more loop. The ASCII we have been talking about is a seven-bit code, but the hexadecimal representation is eight bits. The most significant bit is frequently called the "parity bit," and we have assumed it is set to zero. Some references assume a parity bit of "1". In that case, add \$80 (1000000 binary) to the ASCII values as represented. For example, carriage return becomes \$8D rather than \$0D, and that "N" becomes \$CE rather than \$4E.

One final note before I move on. I said that the dollar sign is used to preface hexadecimal numbers. You might wonder if other bases have similar labels. They do! Binary is denoted by a percent sign, and decimal either by no sign or an ampersand. Thus, %01011010 = \$5A = &90.

Another item in the mailbag this month is from D. A. Scott W4AKQ in Cape Coral, Florida. Scotty is looking for information on "silent RTTY," that is, RTTY using video terminals or computers. Sorry to say, I know of no book devoted exclusively to the subject. More articles have been published here in 73 than just about anywhere on that topic, however, and Fig. 1 is a compilation of articles I have noted within the last several years. Little information on any of the commercial RTTY video terminals has been received by this column, despite promises and offers from several sources. I would encourage you to ask around your area and see what

8080-based systems:

- "A Very Cheap I/O—The Model 15," 73, May, 1976, p. 77.
- "Digital Group RTTY Micro," 73, September, 1977, p. 98.
- "Baudot Interface Cookbook," *Kllobaud*, September, 1978, p. 66.

6502-based systems:

- "Baudot To ASCII," 73, November, 1976, p. 172.
- "RTTY Loop," 73, June and July, 1978 (receiving).
- "RTTY Loop," 73, June and July, 1979 (transmitting).
- "RTTY With The KIM," 73, September, 1977, p. 110.
- "Try Your KIM-1 On RTTY," 73 October, 1977, p. 88.
- "KIM-1 Can Do It," 73, February, 1978, p. 88.
- "RTTY With The KIM," 73, December, 1978, p. 170.
- "RTTY Transceive For The KIM-1," 73, May, 1979, p. 78.

Other systems:

- "Build This Exciting New TVT," 73, March, 1976, p. 76.
- "ASCII/Baudot With A PROM," 73, June, 1976, p. 114.
- "A RTTY/Computer Display Unit," 73, July, 1976, p. 118.
- "The XITEX Video Terminal," 73, December, 1978, p. 132.

Fig. 1. Articles on silent RTTY.

people are using. Be sure to write the manufacturers of any equipment you contemplate buying prior to investing in a unit, for information and literature. There apparently have been some problems with rf sensitivity and interfacing difficulties with some widely advertised equipment.

Another note received in the mail is from Art Santella K1VKO in East Norwalk, Connecticut. Art is using a Model 28-KSR Teletype and has added outboard paper tape equipment to provide the features of the "ASR" set. What he wants to be able to do is punch a reply tape while receiving.

I presume that you have the common Model 14 tape equipment, Art, although this will apply to any standard equipment. The tape punch you have is called a "reperforator," because it punches off of an active TTY line. Data is fed into it via the standard 60 mA [RTTY] Loop (don't mind the plug!). From what you tell me, you are using a HAL ST-6 demodulator as both the source of data and the loop supply. I assume your station hookup is something like Fig. 2.

There are many ways of hooking equipment up together, switches, tie lines, and the like, but the way that many RTTY enthusiasts have come to love over the years is with a patch panel. Diagrammed in Fig. 3, this is a series of phone jacks which are connected in series, so that any equipment plugged

into the strip is in series with anything else. Equipment may thus be switched from loop to loop with a minimum of effort. The key to the system is that the jacks are all shorted if nothing is plugged in. Thus one jack or all may be in the system at any time. Such a plug/jack system may be just what you need.

If you look into your Model 28, you should find a terminal strip, diagrammed in RTTY Loop, October, 1978, page 20, upon which are represented the leads for the keyboard. You should find a black and a brown wire on terminals seven and eight of TB-751. Bring these wires out to a plug and the terminal connections to a shorting jack. When you plug the keyboard leads into this jack, all will be as original. However, when you plug the keyboard into an auxiliary loop, which also supports the reperforator, you will be able to punch tape independent of the printer. Fig. 4 diagrams this system.

If you are going to do this, be sure that your keyboard contacts are clean and bypassed for rf. You normally do not receive while typing, and thus are unaware of the arcing that takes place on many keyboard contacts. If not conditioned, this arcing can generate severe hash in the received signal. A simple loop supply to use to run the auxiliary loop is shown in Fig. 5. Please don't forget that current-limiting resistor. Fried selector magnets do not smell very good.

That's about it for this month. With autumn approaching, conditions should be changing on many of the HF bands, and I hope to be on more and QSO many of you who have written.

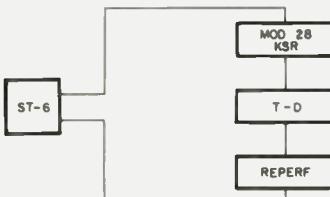


Fig. 2.



Fig. 3. Jack strip wiring.

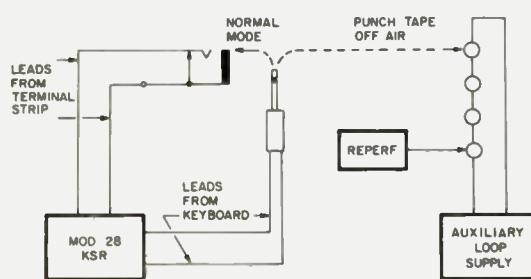


Fig. 4. Sample station wiring.

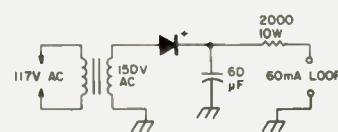


Fig. 5. Simple loop supply.

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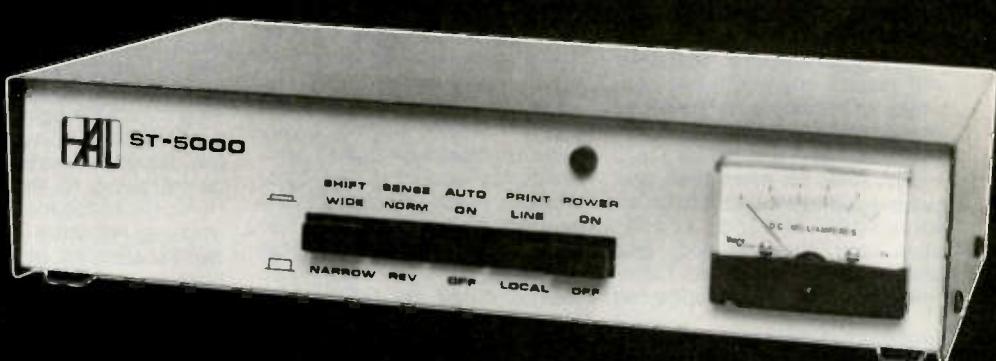
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New Products

SON OF THE QF-1

The recently introduced Autek Research QF-1A active audio filter is a unique selectivity-enhancing accessory that should be of interest to many amateurs, not only because of its quality and features, but also because of its moderate price tag (\$65 at this writing).

The QF-1A is the latest in a line of receiving filters that Autek has marketed since 1972. One could probably trace the lineage of the QF-1A and its competitors to the popular National Radio "Select-o-Ject" audio filter that was the apple of many a ham's eyes in the fifties. The current breed of solid-state filters offers a lot of improvements over the "S-o-J" and its vacuum tubes; the QF-1A, I found, has a number of special features that make it of great interest to CW and SSB operators alike.

Among these features are four filter modes of operation (peak, notch, lowpass, and highpass), continuously variable selectivity or bandwidth from "flat" down to 20 Hz, a main frequency adjustment that can be set anywhere from 250 to 2500 Hz in any of the four modes, and a second notch frequency adjustment control (added since the QF-1—more on this feature later). The filter is housed in a 6½" × 5" × 2½" gray-finish case and is designed to be plugged into the headphone or external speaker jack of the transceiver or receiver with which it is to be used. It contains a well-filtered 115-V ac power supply and a one-Watt audio power amplifier, so it does not require any additional connections to the rig for power or audio tap-in.

I have used the earlier QF-1 model filter for more than two years and have found it to be an essential station accessory—and, I might add, one which has required no maintenance at all. And, despite the

reasonably sharp 600-Hz CW and 2400-Hz SSB i-f filters which my Tempo 2020 has built in as standard equipment, I have found that the QF-1 has made the transceiver's selectivity a great deal sharper. The continuously variable filter characteristics allow a much greater flexibility in combating QRM and in eliciting "armchair copy" in an extremely crowded band.

The original QF-1, written up in *QST* in March, 1977, is similar in design to the newer version but is simpler and has fewer front-panel controls. The QF-1 has selectivity and frequency controls, a three-position mode switch, a speaker/headphone jack, and an ac on/off switch. The QF-1A, apparently drawing on user suggestions and throwing in some "nice to have" features, has five switches and controls on the front panel: (1) an ac on/off switch, which automatically bypasses the filter when turned off; (2) a four-position mode switch; (3) a calibrated selectivity control; (4) a frequency control; and (5) an auxiliary notch frequency control. The headphone/speaker jack is now mounted on the rear panel.

After unpacking the QF-1A from its shipping carton and reading the instructions (which were carefully prepared and complete), I hooked the unit up to my 2020, replacing the faithful QF-1. I found that overall performance was similar, although audio quality was somewhat "cleaner." Like the QF-1, the filter allowed superior "single-signal" reception of very closely spaced CW stations. The filter also worked nicely on SSB, having a very definite impact on signal readability by reducing sideband chatter, static crashes, background noise, and receiver hiss. By using the adjustable notch feature, heterodynes (beats) could be reduced or

eliminated on both CW or SSB with no trouble.

So far, there was little difference between the units. I found, however, that the newer filter provided a few operating features not found in the QF-1. For example, the on/off switch now automatically bypasses the filter when it is turned off. When using the QF-1, you had to unplug the speaker or headphones and plug them back into the transceiver or receiver when you were not actually using the filter, a minor but not insignificant annoyance. The older model's controls were adjustable over a comparable frequency range, but they were not calibrated; the QF-1A's controls are calibrated at key points, a factor which is useful in resetting the filter to the "flat" position when casually tuning the band or in trying to remember "which way to go" with a control in a pileup. The controls also have a somewhat improved "feel" and bandspread. The auxiliary notch control even allows a second heterodyne to be nulled out should the occasion arise (it will!). It's also possible to use this auxiliary feature to combine operating modes so that you can simultaneously notch/peak, notch/lowpass, or notch/highpass to your heart's content; the instructions tell just how to set up the controls to perform all this "black magic."

The new filter, I found, really comes into its own in serious CW work under crowded contest weekend or DXpedition conditions where weak, in-the-grass signals can often be brought up to effective "599" levels even in the presence of close-by rock-crushing signals which would ordinarily obliterate them. The desired signal can be peaked with an effective bandwidth of as little as 20 Hz with minimal ringing (although such narrow bandwidths are not usually necessary and can be fatiguing to the operator). Using the notch mode, very pronounced rejection (up to 70 dB) can be attained with very little tinkering with the controls. I find that operating in the lowpass mode doesn't produce excessive "booming" and is perfectly adequate for casual CW work used in tandem with the Tempo's 600-Hz filter. When the QRM gets rough, I switch to the peak mode and use the selectivity and frequency controls to zero-in on the desired signal. It's interesting to note that even with the transceiver's sharp CW filter, it's possible to actually "tune through" the Tempo's 600-Hz bandwidth and bring up three or four separate CW signals to solid copy that were unreadable without the filter in the circuit.

My use of both filters has indicated that SSB reception is also normally best in the low-pass mode and allows one to easily cope with adjacent-signal splatter and QRM; I like to run with this mode engaged at all times, as it enhances overall audio quality. The auxiliary notch control is effective in nulling out heterodynes, and when the SSB sledding gets rough, all one has to do is to slightly "sharpen up" the selectivity by tweaking the selectivity and frequency controls.

While the QF-1A and other active audio filters work best with receivers and transmitters that have narrow, steep-skirted internal i-f filters, I have found that the filter also does a surprisingly good job on sets that have poor i-f selectivity characteristics. For example, I found that I could dramatically improve the effective selectivity of my Yaesu FRG-7 general coverage receiver. The FRG-7, as supplied from the factory, is much too broad for good SSB and CW work. It's mainly an AM-type SWL receiver, and a good one for that purpose. Its basic i-f selectivity is listed as ±3 kHz at 6 dB down, and a built-in fixed audio filter allows selection of either 3-kHz, 2500-Hz, or 1500-Hz bandwidths. When hooked into the QF-1A, I could tell little practical difference in overall CW and SSB selectivity characteristics between the FRG-7 and the Tempo 2020 with its excellent internal filters! CW signals which were completely unreadable on the barefoot FRG-7 were now a solid Q-5 copy with the filter switched in, and the improvement in SSB reception, though not quite as dramatic, was nevertheless impressive. Of interest to SWLs, I found that the filter also improved AM selectivity on both the shortwave and AM broadcast bands and allowed stations just a few kHz apart to be separated nicely with little cross-channel chatter. This ability would be of special help to the medium-wave DXer in trying to sort out split-channel AM broadcast stations (those operating between the standard 10 kHz U.S. channels), as well as foreign broadcast stations which are often but 3 or 4 kHz apart on the shortwave bands. (If you use the QF-1 with the FRG-7 or another receiver with a built-in fixed audio filter, run with the internal filter out for best results.)

Being an inquisitive sort, I tried out the new filter with some other equipment and was surprised by the results. For example, I used it quite effectively with a converted Johnson 4740 SSB transceiver on ten meters. When operated in the



Autek's QF-1A.

lowpass mode, the filter did a great deal to reduce channel "bleedover" effects and crosstalk. Also, the notch feature was handy in suppressing annoying heterodynes from the many AM stations now on the band that use converted AM CB transceivers.

I also experimented with the filter on my KLM-2800. The 2700 is designed for two meter all-mode work, though the primary emphasis—as in most other "all-mode" transceivers—is on FM; it is somewhat lacking in good CW selectivity, and there are no provisions to add optional i-f filters to reduce bandwidth. Hooking up the QF-1A had much the same beneficial effect as it did when it was worked with the Tempo 2020—only more so. The filter added a great deal of selectivity to an otherwise broad set for serious CW work, and it made a difference on SSB as well. As expected, it didn't do much for FM reception.

The original QF-1 design was a good one. Most of the QF-1A's changes are convenience features and minor improvements, such as the addition of the automatic bypass, the auxiliary notch control, and the calibration marks on the selectivity/frequency controls. Also, the newer version's selectivity goes down to 20 Hz, as opposed to a "mere" 50 Hz on its predecessor. Both designs are thoroughly bypassed and filtered to prevent rectified rf from your transmitted signal from getting back into the audio circuitry and causing problems; I had no difficulty from rf feedback even when running a full kilowatt on 80 through 10 meters.

When working with very narrow selectivity, you may find it convenient to connect a Y-cord adapter in the output line and run an output to an oscilloscope or ham-type monitor scope (such as the Heath SB-610 or similar Yaesu and Kenwood versions) to visually help in determining exact audio peaking. Also, one feature that could have been added to the QF-1A but wasn't is a set of switches to allow its use with two audio sources (two receivers, or a transceiver and a receiver) and two outputs (a headphone and a speaker). As it stands now, the unit must be manually plugged in and out of the equipment with which it is to be used, and the speaker and headphones must be plugged and unplugged by hand.

Also, I have found it expedient to feed the filter's output to a small "wireless FM rebroadcaster" module (I use the Ramsey module @ \$3) used in conjunction with a pair of cordless FM headphones. Doing so

eliminates the need for clumsy headphone coiled cords and has the side benefit of allowing you to monitor the band anywhere in your home or yard by tuning in the rebroadcaster's signal on a nearby FM radio. You may not want to try this, but I mention it to point out the possibilities and advantages of hands-off, cord-free hamming.

Some suggestions I had for enhancing the versatility of the basic QF-1 were described in my May, 1978, *73 Magazine* article, "The Super Select-o-Ject." Many of the changes I made to the QF-1, such as the switching arrangement, can be adapted to the new filter.

One caution: The QF-1 and its successor tend to emphasize any residual ac hum present in the transceiver's or receiver's audio output. This is normally caused by insufficient ac filtering in the rig and not in the audio filter. I have found this condition to be noticeable mainly when operating in the lowpass mode and feeding low-impedance, hi-fi type headphones, which normally have a heavy bass response. This problem can be minimized by adding a 50-to-150-Ohm, ½-Watt resistor in series with the headphone lead to cut down the phones' sensitivity and low frequency response.

In my opinion, the QF-1A is a real jewel well worth the modest price tag. While I have not as yet made an in-shack comparison between the Autek filter and its competition, I can say that the quality of construction (steel case) and circuitry board is excellent and reflects care in design and manufacture. The QF-1A makes a modest receiver or transceiver "come alive" in the selectivity department and allows the average set—so often lacking in *real* CW performance—to become a winner. For further information, contact Autek Research, Box 5127, Sherman Oaks CA 91403. Reader Service number A100.

**Karl T. Thurber, Jr. W8FX/4
 Ft. Walton Beach FL**

Notes

1. There are a large number of audio communications filters on the market today. These include some highly competitive designs by MFJ, Waneco Radio, Dynamic Electronics, Datong, Kantronics, Electronic Research Corporation of Virginia, and others. These filters range in price from about \$30 to \$200 and more. If you are interested in building your own, a very good active filter design is shown by Donald Morar W3QVZ in the December, 1978, issue of *73*. His article, "Build the Flexi-Filter—A Very Active Device," is on page 222.

References

1. Product review, Autek Research QF-1 R-C Active Filter, *QST*, March, 1977, p. 44.
2. Thurber, Karl T., Jr. W8FX/4, "The Super Select-o-Ject," *73 Magazine*, May, 1978, p. 116.



Non-Linear System's MS-215 Miniscope.

MISSION IMPOSSIBLE? NOT WITH THE MS-215!

The manufacturer's service information reminded me of that TV series "Mission Impossible," and for that matter, what I had to do could very well have been a script for the show. It was nothing exciting—I just had to identify one bad chip on a circuit board containing 36 chips, located in a cabinet with an access door about 18 inches square. But, not to worry, according to the manufacturer—just get into the cabinet with a dual-trace 10-MHz scope, triggered, and look for the inputs and outputs on each chip. When you don't find both conditions, you've found the bad chip. I did omit one minor consideration—the board was a part of a computer-based central station monitor at our hospital, was hardwired in, and had to remain operational during the entire test procedure.

Not to be thwarted, I reached into my tool set and removed just what the "doctor" ordered, the Non-Linear System's MS-215 15-MHz dual-trace scope. Besides being self-contained and self-powered by nicads, the Miniscope measures a mere 2.9" x 6.4" x 8.0". Its vertical bandwidth extends from dc to 15 MHz, with ac, dc, or ground coupling switch-selectable. Deflection is 10 mV/div to 50 V/div in 12 calibrated ranges (with a stated accuracy of 3% of full scale with the vernier in the full clockwise position).

The horizontal characteristics include: internal timebase or external horizontal switch-selectable mode with a timebase of 0.1 usec/div to 0.5 sec/div in 21 calibrated ranges. The maximum input voltage for both horizontal and vertical sections is 350 V dc and peak ac, providing that the dc component does not exceed 250 volts. A switch-selectable trigger permits + or -, with modes of internal, external, and line (line is non-functional when operating

off the nicads). A trigger-level control permits continuous adjustment of the trigger point.

While all of these technical specifications will impress a number of you, why should you or any ham want a scope this tiny? The answer is quite simple: The MS-215 Miniscope can go to the work when the work can't go to the scope! Maybe that's an oversimplification of the issue, but how many times in the last six months have you wished that you could have a piece of test equipment which was scaled down in size to the IC circuitry being examined, be it a computer CPU or the innards of a synthesized 2-meter ham rig? Better yet, how many jobs could you have done with little or no sweat if you and your scope could have been right there—on the spot rather than tied to a large scope with equally large probes further hindered by an ac power cord.

I acquired an MS-215 because with it we could locate the defective chip and remove and replace it at a total cost of perhaps \$2.00 in parts, while the new board would have cost somewhere in the area of \$1200. Considering that the MS-215 sells for around \$400 from a number of advertisers in *73*, purchasing it was a very wise move in terms of dollars and cents!

What can't you do with the MS-215? Well, to be honest, very little—unless you want to display a waveform to more than two close friends. We have literally used the scope for everything and anything you could imagine—from probing the interior works of a piece of hospital gear to troubleshooting the somewhat sickly departmental HT. In all cases, it performed as well as one of the larger and considerably higher-priced scopes in the department. Considering that the small package even included a calibrator which provides a square-wave signal of 1 volt peak-to-peak, and a graticule



Nye Viking's Model 250-0046-001 phone patch.

which (while small like the CRT) is easy to read — what more can you ask for in a package you can comfortably stuff into a coat pocket?

If you are interested in what you've read thus far and want more information about the MS-215 Miniscope, drop Mr. Allan Kay a line at *Non-Linear Systems, Inc., Dept. 73, Box N, Del Mar CA 92014*. Be sure to tell him that you read about their products in *73 Magazine*. Reader Service number N22.

**Elliot S. Kanter W4PGI
Savannah GA**

**NYE VIKING RECEIVES
FCC REGISTRATION
FOR PHONE PATCH**

Approval by the Federal Com-

munications Commission for official registration under Part 68 of the FCC Regulations has been granted to the Wm. M. Nye Company, Bellevue WA. This allows the Nye Viking phone patch to be plugged directly into the telephone line without the need (or cost) of a telephone company-supplied coupling device. It should be noted that users must still notify the telephone company that they are connecting the phone patch to the telephone line, and must furnish the company the official registration number and the ringer equivalence number (which are printed on an attached label).

Telephone patches may not legally be connected to party lines or pay telephone lines.

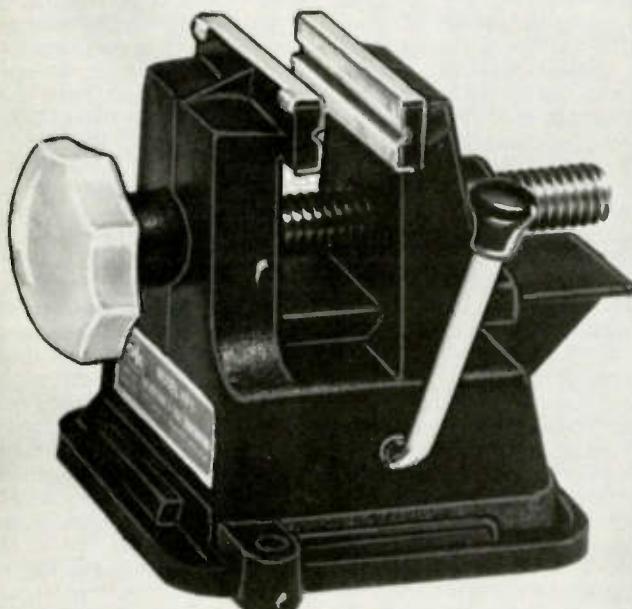
Users are cautioned that they must comply with all other requirements of the FCC pertaining to amateur radio communications.

The Nye Viking phone patch comes in two models: the 250-0046-001, without speaker, which provides connection to your own external speaker, and the 250-0046-003, with built-in loudspeaker, for use with most transceiver installations.

Nye Viking phone patches manufactured prior to official FCC approval and registration can be upgraded to approved status with the necessary changes, which include a 7-foot cord and a plug to connect into the telephone company line socket. Units returned for modification should be carefully packed and contain the sender's name and address. For further information, contact the Wm. M. Nye Company, Inc., 1614 130th Ave. NE, Bellevue WA 98005. Reader Service number N4.

LIGHT-DUTY VISE

The new Model VV-1 from OK Machine and Tool Corporation is a unique vacuum-based light-duty vise for precision handling of small components and assemblies. Featuring rugged ABS construction, it has 1½" wide jaws and 1¼" travel for maximum versatility. Also featured are an oversize knob for precise positioning and screw lugs for permanent installation. For further information, contact OK Machine and Tool Corporation, 3455 Conner Street, Bronx NY 10475. Reader Service number O5.



OK Machine and Tool's new VV-1 vise.

sturdy structure capable of supporting considerable antenna arrays. The modularity of the six-foot sections makes them a natural for site surveys, field operations, and portable communications tests of all sorts, including amateur radio EME (moonbounce) operations.

Taking advantage of a building's height, they readily mount on rooftops, flat or peaked. The use of a length of 2 x 4 under leg pairs provides a simple yet effective mount for the tower. Since it must be guyed, no bolts, nails, or other requirements are necessary. The 2 x 4s additionally serve to distribute the weight over several roof joists.

These portable units are readily UPS shipped, and may also be carried as luggage on any airline. The construction is not unlike a childhood toy Erector set, with each module bolted together with supplied hardware. Optional stainless steel hardware is also available (S suffix). Rotor and thrust bearing mounting plates are included in the basic tower package (Model LT-1), which yields an 11-foot structure when erected. The 6-foot add-on modular sections (Model LT-2) increase the height up to nearly 30 feet.

Since the tower is built from angle pieces, it forms its own ladder when erected and guyed. The base is 43 inches square, and the tower sections are 9.5 inches square. For further information, contact *Lunar Electronics, 2785 Kurtz Street, Suite 10, San Diego CA 92110; (714)-299-9740*. Reader Service number L17.

**LUNAR'S MODEL LT-1
MODULAR TOWER**

Lunar Electronics has announced a new line of towers. Manufactured in Lunar's San Diego plant, these towers are all-aluminum angle pieces, which bolt together to form a

INSTA-PAC

Something new has come upon the scene which promises ease and efficiency for the QSL fan. Its name is "Insta-Pac," and it affords the ham the oppor-

Continued on page 40



Lunar's LT-1 modular tower.

This NEW MFJ Versa Tuner II . . .

has SWR and dual range wattmeter, antenna switch, efficient airwound inductor, built in balun. Up to 300 watts RF output. Matches everything from 1.8 thru 30 MHz: dipoles, inverted vees, random wires, verticals, mobile whips, beams, balanced lines, coax lines.

MFJ LOWER PRICES!



Transmitter matching capacitor. 208 pf. 1000 volt spacing.

Sets power range, 300 and 30 watts. Pull for SWR.

Meter reads SWR and RF watts in 2 ranges.

NEW, IMPROVED MFJ-941B HAS . . .

- More inductance for wider matching range
- More flexible antenna switch
- More sensitive meter for SWR measurements down to 5 watts output

NEW LOWER PRICE

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Antenna matching capacitor. 208 pf. 1000 volt spacing.

Only MFJ gives you this MFJ-941B Versa Tuner II with all these features at this price:

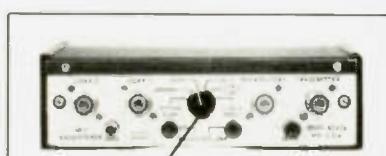
A SWR and dual range wattmeter (300 and 30 watts full scale) lets you measure RF power output for simplified tuning.

An antenna switch lets you select 2 coax lines direct or thru tuner, random wire/balanced line, and tuner bypass for dummy load.

A new efficient airwound inductor (12 positions) gives you less losses than a tapped toroid for more watts out.

A 1:4 balun for balanced lines. 1000 volt capacitor spacing. Mounting brackets for mobile installations (not shown).

With the NEW MFJ Versa Tuner II you can run your full transceiver power output — up to 300 watts RF power output — and match your



ANTENNA SWITCH lets you select 2 coax lines direct or thru tuner, wire/balanced line, dummy load.

transmitter to any feedline from 160 thru 10 Meters whether you have coax cable, balanced line, or random wire.

You can tune out the SWR on your dipole, inverted vee, random wire, vertical, mobile whip, beam, quad, or whatever you have.

You can even operate all bands with just

one existing antenna. No need to put up separate antennas for each band.

Increase the usable bandwidth of your mobile whip by tuning out the SWR from inside your car. Works great with all solid state rigs (like the Atlas) and with all tube type rigs.

It travels well, too. Its ultra compact size 8x2x6 inches fits easily in a small corner of your suitcase.

This beautiful little tuner is housed in a deluxe eggshell white Ten-Tec enclosure with walnut grain sides.

SO-239 coax connectors are provided for transmitter input and coax fed antennas. Quality five way binding posts are used for the balanced line inputs (2), random wire input (1), and ground (1).

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Same as MFJ-941B but less 6 position antenna switch.

NEW MFJ-944 HAS 6 POSITION ANTENNA SWITCH ON FRONT PANEL. NEW LOWER PRICE

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Same as MFJ-941B but less SWR/Wattmeter.

NEW MFJ-943 MATCHES ALMOST ANYTHING FROM 1.8 THRU 30 MHZ. NEW LOWER PRICE

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Same as MFJ-941B, less SWR/Wattmeter, antenna switch, mounting bracket. 7x2x6 in.

ULTRA COMPACT 200 WATT VERSA TUNERS FOR ALL YOUR NEEDS.

MFJ-901 VERSA TUNER MATCHES ANYTHING, 1.8 THRU 30 MHZ. NEW LOWER PRICE

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Efficient 12 position air inductor for more watts out. Matches dipoles, vees, random wires, verticals, mobile whips, beams, balanced lines, coax. 200 watts RF, 1:4 balun, 5x2x6 in.

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Same as MFJ-901 but less balun for balanced lines. Tunes coax lines and random lines.

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Microcomputer Interfacing

Peter R. Rony
 Jonathan A. Titus
 Christopher A. Titus
 David G. Larsen

MULTIPLEXED LED DISPLAY

Regardless of their application, most microcomputers need peripheral I/O devices for the input and output of data. The more common output devices included seven-segment displays, 5 x 7 dot matrix displays, teletypewriters, and cathode-ray tube (CRT) displays. We shall discuss several methods that can be used to interface seven-segment displays, as well as several different programs that are typically used to "drive" such displays, which are widely used in electronic games, calculators, point-of-sale (POS) terminals, gasoline pumps, children's toys, and taxi meters.

One of the simplest methods that can be used to interface a seven-segment display to a microcomputer consists of latching the appropriate data values from the data bus under software control (Fig. 1). The latch's inputs (7475) are wired to the

microcomputer's bidirectional data bus and the latch's outputs are wired to a seven-segment decoder/driver (7447). The decoder's outputs (current sinking) are wired to the seven-segment display with 220-Ohm current-limiting resistors. When an OUT 125 instruction is executed, the content of the 8080's A register is latched by the display interface and the two BCD numbers represented by D7-D4 and D3-D0 are illuminated on the two displays. The instructions listed in Fig. 2 cause a 39 to be displayed.

To display a 10-digit number using this method, ten latches, ten decoder/drivers, seventy resistors, and ten seven-segment displays will be required. One method of reducing the "parts count" for this interface would be to use a device such as the 8255 Programmable Peripheral Interface (PPI) integrated circuit.¹ This device can be used as three independent eight-bit output ports, so it is the equivalent of six 7475 latches. Therefore, two 8255 PPI chips, along with 10 decoders/drivers, seventy resistors,

and ten seven-segment displays would be required in the interface. One obvious disadvantage of this interfacing method is the large number of integrated circuits required. However, one advantage is that the software required to drive this interface is relatively simple. Also, the microcomputer only has to output this information once to the interface for the information to be continuously displayed. This, of course, is due to the latches or the 8255 chips in the interface. Thus, the microcomputer can output numeric information once and then go on to perform any other required operations.

Another interfacing method that can be used is digit multiplexing. Multiplexing reduces the display interface electronics (number of parts) to a minimum, but at the expense of longer and more complex display driver software. Multiplexing a display consists of enabling or turning on one particular digit with a *digit enable code* and providing the BCD numeric information for that digit to a multi-digit display interface. In this way, each digit is turned on, one at a time, as the actual BCD data for each digit is provided. Multiplexing is usually only used with multi-digit displays. As an example, suppose that the number 237 is to be displayed on a three-digit multiplexed display. To display this number, the BCD value for the digit seven would be output to the interface, along with the digit enable code for the right-hand display. After a short period (1 to 10 ms), the BCD value for the three would be output, along with the digit enable code for the middle digit. Again, after a short delay, the BCD

value for the two and the digit enable code for the left-hand display would be output to the interface. By performing this sequence fifty or more times every second, each digit in the display appears to be on all of the time. This same display method is used in hand-held calculators. Even though the digits are being turned on and off, it is happening too fast for the eye to see. The interface for a 10-digit multiplexed display is shown in Fig. 3.

When an OUT 125 instruction is executed, bits D3 through D0 of the A register will determine which one of the ten digits in the display will be enabled (turned on). Therefore, these four bits constitute the digit enable code. Bits D3 through D0 are latched (7475) and are decoded with a one-of-ten decoder (7442). The decoded outputs of the 7442 are wired to the common cathodes of the individual digits in the display. Bits D7 through D4 will provide the BCD code of the value to be displayed (0 through 9). These bits are also latched (7475) and are decoded by a seven-segment decoder/driver (DS8857, National Semiconductor Corporation, Santa Clara CA). The DS8857 supplies the current required to turn on the various segments (A-G) within the enabled digit selected by the 7442 decoder chip.

A relatively simple program can be written in which five packed BCD words (two BCD digits per word) are output to the display so that a 10-digit number is displayed (Fig. 4). This program has to unpack the BCD words stored in memory, combine the BCD digits with a

Continued on page 150

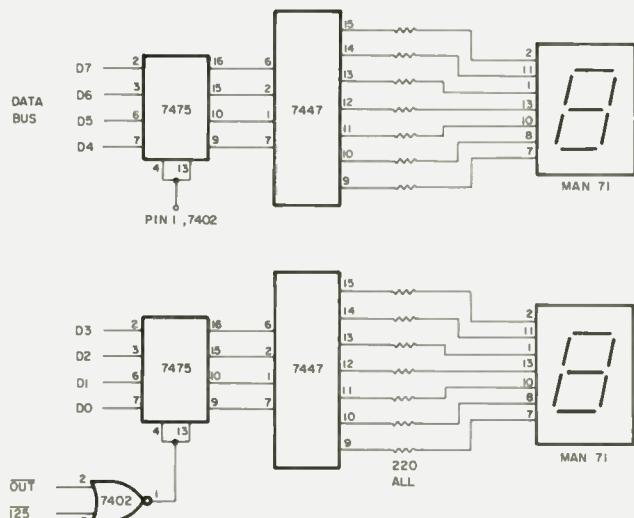


Fig. 1. A simple two-digit LED display interface.

```
/THIS SECTION OF A PROGRAM OUTPUTS THE
/8-BIT PATTERN 00111001 (OCTAL 071, HEX 39)
/TO AN OUTPUT PORT EQUIPPED WITH TWO
/SEVEN-SEGMENT DISPLAYS.

.
.
.
MVI A   /LOAD A WITH THE FOLLOWING IMMEDIATE
071   /DATA BYTE (HEX 39, BINARY 00111001)
OUT   /OUTPUT IT TO THE TWO SEVEN-SEGMENT
125   /DISPLAYS EQUIPPED WITH LATCHES
.
.
.
/CONTINUE WITH THE REMAINDER OF THE
/PROGRAM
```

Fig. 2. Displaying a 39 on the two-digit display.

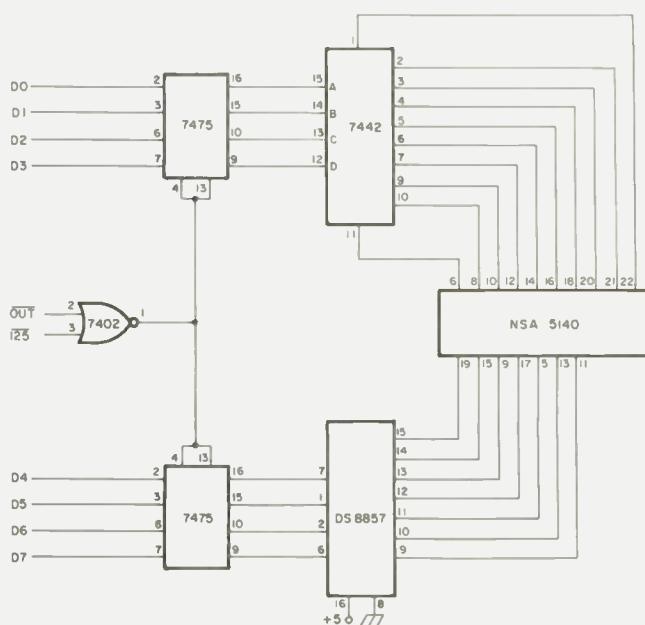
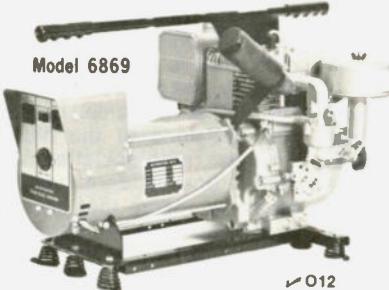


Fig. 3. A 10-digit multiplexed LED display interface.

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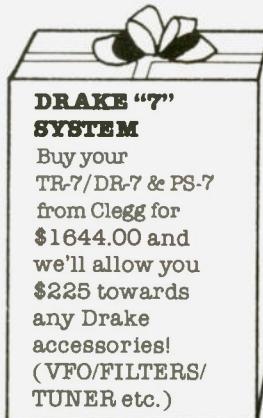
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Awards

Bill Gosney WB7BFK
2665 North 1250 East
Whidbey Island
Oak Harbor WA 98277

After months of planning and careful consideration, I'm particularly proud to announce this 73 exclusive—the 73 Magazine Awards Portfolio. Consisting of four domestic award incentives and four DX achievement programs, the awards portfolio promises to be a challenge and capture the interest of almost everyone on the band, whether you are a casual rag chewer or a big-time contestor.

Read through the various award rules with caution. The requirements are not as easy as one might first imagine. We want our award recipients to know they had to earn their recognition and therefore designed each award to be somewhat of a challenge.

I will present each award in detail—the DX-oriented awards will be in the spotlight this month, and next month I will focus on the domestic ones.

Each of the sponsored awards will offer its own degree of difficulty. None was designed to be an overnight accomplishment, nor were any meant to duplicate any other awards in existence today.

So tune up those QRP rigs, kick in the blower on the kilowatt, rotate those beams, and fine-tune that dipole, as time's a wastin'; you've got some contacts to make and some awards to win! I plan to feature a special profile of the winner of the first award issued in each category. Subsequent recipients will be recognized in a monthly award summary.

73 DX COUNTRY CLUB AWARD

1. Sponsored by the editors of 73 Magazine.
2. Available to licensed amateurs throughout the world.

3. To be valid, all contacts claimed must be made in a single calendar year (January 1 through December 31), beginning January 1, 1979, and after.

4. This award is available for all phone, CW, and mixed modes.

5. To qualify, a minimum of 73 DX countries must be worked and confirmed from the 73 Magazine WTW (Work the World) DX Listing. All contacts must be made in the same calendar year.

6. Annual endorsement stickers are available for each succeeding year in which a minimum of 73 DX countries are worked.

7. To apply, prepare a list of

claimed contacts in prefix order. Include each station's full callsign, date and time in GMT, mode, and band of operation:

8. Do not send QSL cards! Have your list of contacts verified by two local amateurs, a local club secretary, or a notary public.

9. Enclose your verified list and award fee of \$3.00 or 8 IRCs for each award or endorsement. Send to: Bill Gosney WB7BFK, 73 Awards Editor, 2665 North 1250 East, Oak Harbor, Whidbey Island, Washington 98277 USA.

DX CAPITALS OF THE WORLD

1. Sponsored by the editors of 73 Magazine.

2. Available to licensed amateurs throughout the world.

3. To be valid, all claimed contacts must be made January 1, 1979, or after. There are no mode or band restrictions.

4. To qualify, applicants must work and confirm fifty (50) different national capital cities located in DX countries shown on the WTW (Work the World) DX Listing.

5. To apply, make a self-prepared list of contacts made in prefix order indicating the station worked, date and time in GMT, band of operation, and the name of the capital city and DX country.

6. Do not send QSL cards! Have your list of contacts verified by two amateurs, a local club secretary, or a notary public.

7. Enclose your application list and award fee of \$3.00 or 8 IRCs to: Bill Gosney WB7BFK, 73 Awards Editor, 2665 North 1250 East, Oak Harbor, Whidbey Island, Washington 98277 USA.

TEN METER DX DECADE AWARD

1. Sponsored by the editors of 73 Magazine.

2. Available to licensed amateurs worldwide.

3. All contacts must be made on the 10 meter band using only channelized converted Citizens Band equipment or similar type commercial units operating a maximum of 15 Watts PEP output. External amplifiers may not be used.

4. To be eligible for award credit, all contacts must be made October 1, 1978, or after, on AM, SSB, CW, or FM. Mixed mode contacts are not valid.

5. To qualify, applicant must work and confirm at least ten (10) DX countries from the WTW (Work the World) DX Listing. Endorsements will be given for 25, 50, 75, and 100 countries confirmed.

6. To apply, make a self-prepared list of contacts claimed, giving the callsign of each station worked in prefix order. Include the date and time in GMT, band, mode, and a brief description of the equipment used in making each contact.

7. Do not send QSL cards! Have your list verified by two amateurs, a local club secretary, or a notary public.

8. Forward your application list and award fee of \$3.00 or 8 IRCs to: Bill Gosney WB7BFK, 73 Awards Editor, 2665 North 1250 East, Oak Harbor, Whidbey Island, Washington 98277 USA.

WORK THE WORLD AWARD

To enhance the enjoyment of working DX, the editors of 73 Magazine take special pleasure in introducing what we feel will be one of the most sought-after awards in existence today—the Work the World Award.

1. Sponsored by the editors of 73 Magazine.

2. Available to licensed amateurs throughout the world.

3. To be valid, all contacts must be made January 1, 1979, or after. There are no band or mode restrictions. Only DX countries shown on the WTW (Work the World) DX Listing qualify.

4. The ultimate goal of this award program is the Work the World (WTW) Award, signifying excellence in DX communications. However, to qualify for WTW, applicants must first meet the requirements of the six Continental Awards, each of which is a worthy goal in itself.

5. Requirements for Continental Awards: North American Awards—work 13 North American countries; South American Award—work 12 South American countries; European Award—work 12 European countries; African Award—work 12 African countries; Asiatic Award—work 12 Asiatic countries; Oceanic Award—work 12 Oceanic countries.

6. The Work the World Award is issued at no cost to any applicant who meets the requirements of all six Continental Awards. The operator who earns WTW has truly "worked the world."

7. To apply for any of the six Continental Awards, prepare a separate list of claimed contacts for each continent, listing all callsigns in prefix order. Include date, time in GMT, and band.

8. Do not send QSL cards! Have your list(s) verified by two amateurs, a local club secretary, or a notary public.

9. Enclose your verified list(s) and award fee of \$3.00 or 8 IRCs for each award applied for. Forward to: Bill Gosney WB7BFK, 73 Awards Editor, 2665 North 1250 East, Oak Harbor, Whidbey

Island, Washington 98277 USA.

Throughout this column, reference has been made to the new WTW (Work the World) DX Listing. In order to avoid the impossible task of having to decide what is and what is not a country, the awards editor and the staff at 73 decided to accept the decisions of the national amateur radio societies of the world. Special care has been taken to ensure that no single organization will dictate the criteria used; instead, an accumulative effort by all concerned will be recognized.

Thus, as the basis for all the DX awards sponsored by 73 Magazine, the official WTW DX Listing will be categorized by continents of the world. Within each continent, DX countries will appear in order of their call-sign prefix.

Additional copies of the WTW DX Listing or rules of the various award programs are available on request. Please be sure to enclose a 4½" x 9" SASE with all inquiries. Forward your inquiry to my home address which appears throughout this column.

If DX isn't your cup of tea, then the domestic awards featured next month will interest you for sure.

Keep the cards and letters coming. Your comments concerning this column have been well received and appreciated. Should you learn of an awards program being featured in your city, state, province, or country, please forward this information on to me. Together we can share the many achievements being offered our readers throughout the world.

WTW DX LISTING

NORTH AMERICA

C6	Bahamas
CO	Cuba
FG	Guadeloupe
FG, FS	Saint Martin
FM	Martinique
FO	Clipperton Is.
FP	St. Pierre & Miquelon
HH	Haiti
HI	Dominican Republic
J3, VP2G	Grenada & Dependencies
KC4, KP1	Navassa Is.
KG4	Guantanamo Bay
KL7	Alaska
KP4	Desotoe
KP4	Puerto Rico
KS4, KP3, HK0	Serrana Bank and Roncador Cay
KV, KP2	Virgin Islands
OX, XP	Greenland
PJ6, 8	Saba Is.
VE	Canada
VE1	Sable Is.
VE1	St. Paul Is.
VO	Newfoundland, Labrador
VP2A	Antigua, Barbuda
VP2D	Dominica
VP2E	Anguilla
VP2K	St. Kitts
VP2L	St. Lucia
VP2M	Montserrat
VP2S	St. Vincent & Dependencies
VP2V	British Virgin Islands
VP5	Turks and Caicos Islands
VP9	Bermuda
W, K. N. A.	United States of America
XE	Mexico
XF4	Revillagigedo Islands
ZF	Grand Cayman Islands
4U	HQ, United Nations
8P	Barbados

Continued on page 155

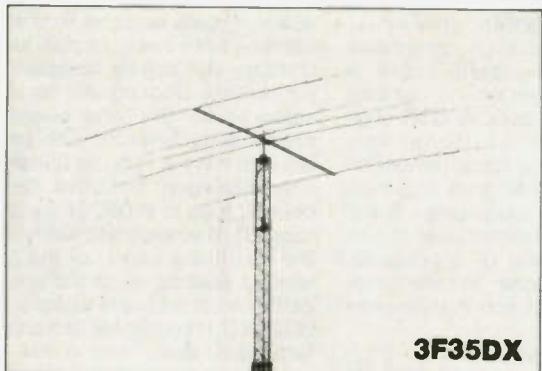
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ANTENNA SYSTEMS

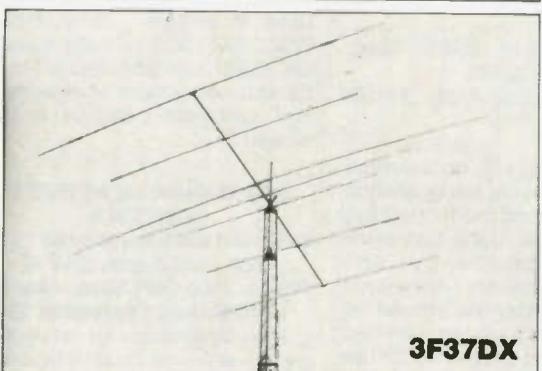
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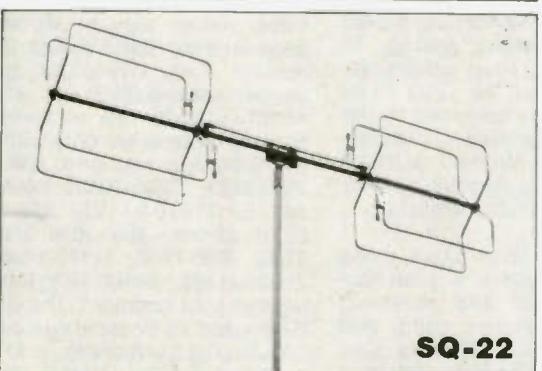
These beams employ hybrid system which is a combination of separated full-size driven element for each band individually and Hi-Q trap parasitic elements. These feature result high radiation efficiency, high power rating and excellent VSWR in entire band width.



3F35DX



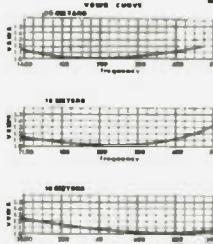
3F37DX



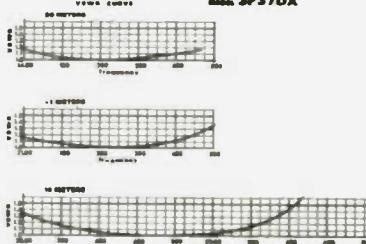
SQ-22

MODEL	3F37DX	3F35DX
BAND	14 21 28	14 21 28
ELEMENTS	7	5
ELEMENTS PER BAND	20m 3 15m 5 10m 5	20m 3 15m 5 10m 3
ANTENNA GAIN	20m 8.5dB 15m 10dB 10m 10dB	20m 8.0dB 15m 8.5dB 10m 8.0dB
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VSWR	1.5:1 BETTER	1.5:1 BETTER
IMPEDANCE	50Ω	50Ω
MAX. ELEMENT L.	10.5m	10.5m
BOOM LENGTH	7.5m	5.0m
BOOM DIAMETER	50mm	50mm
TURNING RADIUS	5.3m	5.25m
WIND RATING	40m/sec.	40m/sec.
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MULTI BAND BEAM DX SERIES 3F37DX



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Contests

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Please note that with this issue all award and contest announcements are being split into their respective separate columns. Please forward all information directly to the individual responsible for the particular column. Any information sent via 73 Magazine headquarters in Peterborough NH is only delayed and may not be received by the appropriate editor in time to appear in the magazine. Also, to the organizations sponsoring a contest, don't forget to send an abbreviated copy of the results for possible publication. They will be printed as space permits.

KENTUCKY QSO PARTY

Starts: 0001 GMT September 15
Ends: 2359 GMT September 16

This contest is sponsored by the Bluegrass Amateur Radio Club. Only one contact per band or mode with each station. No repeater or pre-arranged contacts are allowed. Suggested frequencies are lower edge of General and Novice bands, 2 meter simplex and SSB, and 6 meters.

EXCHANGE:

RS(T), county, and consecu-

tive QSO number; non-KY stations substitute state, country, or province for county.

SCORING:

Score 1 point per QSO. KY stations multiply by number of states, countries, and provinces worked. Non-KY stations use number of KY counties worked. Final multipliers are: 1.5 for all 10 meter contacts between 0200 and 1000 GMT; 2.0 for input of 15 Watts or less; 1.5 for input of 200 Watts or less; 2.0 for all VHF simplex QSOs. Novices or Technicians may take an additional 2.0 multiplier for CW only.

ENTRIES:

Mail logs by October 5 to: Donald Page WD4HPL, 309 Pocahontas Trail, Georgetown KY 40324.

WASHINGTON STATE QSO PARTY

Contest Periods:

0100 to 0700 GMT September 15
1300 GMT September 15 to 0700 GMT September 16
1300 GMT September 16 to 0100 GMT September 17

The 14th annual Washington State QSO Party sponsored by the Boeing Employees' Amateur Radio Society (BEARS) is open to all amateurs. All bands and modes may be used. Stations may be worked once on each band and mode for contact

points and more than once each band/mode if they are additional multipliers.

EXCHANGE:

QSO number, RS(T), county or state/province/country.

SCORING:

Washington stations score two points for each phone contact and 3 points for each CW contact, including contacts with other WA stations. Then multiply by the total of different states, Canadian provinces, and other foreign countries worked. All others score 2 points for each phone contact and 3 points per CW QSO with Washington stations and multiply by the total number of different Washington counties worked (39 maximum). There will be an extra multiplier of one for each group of 8 contacts with the same Washington county for all non-Washington stations.

FREQUENCIES:

CW – 1805, 3560, 7060, 14060, 21060, 28160.
Phone – 1815, 3925, 7260, 14305, 21380, 28580.

Novice – 3725, 7125, 21150, 28160.

AWARDS:

Certificates will be awarded to the highest-scoring station (both single- and multi-operator) in each state, Canadian province, foreign country, and Washington county. Additional certificates may be issued at the discretion of the Contest Committee. Worked Five BEARS Awards are also available to anyone working 5 club members before, during, or after the QSO Party unless previously issued. All QSO Party entries will be screened by the Contest Committee for possible awards. The Worked 3 BEAR Cubs Award is also available for working 3 Novice members.

ENTRIES:

Logs must show dates, times in GMT, stations worked, exchanges sent and received, bands and modes used, and scores claimed. Include a dupe sheet for entries with more than 100 QSOs. Each entry must include a signed statement that the decision of the Contest Committee will be accepted as final. No logs can be returned. Results of the QSO Party will be mailed to all entrants; an SASE is not required. Log sheets and summary sheets must be postmarked no later than October 17 and sent to: Boeing Employees' Amateur Radio Society, c/o Contest Committee, Willis D. Propst K7RS, 18415 38th Ave. S., Seattle WA 98188.

89'ers RUN

Contest Periods:

Run #1: Saturday,

September 15,

1400 to 1800 GMT

Run #2: Saturday,

September 22, 2100 to 0100 (September 23) GMT

The object of this QSO party sponsored by the Sooner Chapter of Ten-Ten International is to collect as many 89'ers' numbers as possible and you do not have to have an 89'er number to enter. All contestants, other than local and charter members, may operate *only one* of the two four-hour runs listed above. Locals and charters will operate both runs. Locals and charters will not be competing for awards. Scoring will be the same as for the 89'er awards, except that contacts with people who have a Ten-Ten number only shall count one point. Send contest logs to WD5CSK by October 31. If you wish to apply for the certificate or any of the advanced awards, separate application must be made directly to WB5TKD. If you would like an information sheet and a list of awards, send an SASE to contest manager Ron Reid WD5CSK, 8832 NW 80, Yukon OK 73099. Contest logs will not be returned unless an envelope and sufficient postage is enclosed.

SCANDINAVIAN ACTIVITY CONTESTS

CW: 1500 GMT September 15 to 1800 GMT September 16

Phone: 1500 GMT September 22 to 1800 GMT September 23

Non-Scandinavian stations try to work as many Scandinavian stations as possible. The same station may be worked once on each band during the contest. Only CW-to-CW and phone-to-phone QSOs are valid, no cross-mode. Use all bands, 80 to 10 meters, but only within the following sections: CW – 3505-3575, 7005-7040, 14010-14075, 21010-21125, 28010-28125; phone – 3600-3650, 3700-3790, 7050-7100, 14150-14300, 21200-21350, 28400-28700 (as legal in your country!). The prefixes used in Scandinavia are: LA/LB/LG/LJ – Norway; JW – Svalbard and Bear Island; JX – Jan Mayen; OF/OG/OH/OI – Finland; OH0 – Aaland Island; OJ0 – Market Reef; OX – Greenland; OY – Faroe Island; OZ – Denmark; SJ/SK/SL/SM – Sweden. All these prefixes are geographically not in Scandinavia, but they are considered so for the contest. Operator classes include: a) single operator; b) multi-op/single transmitter; c) multi-op/multi-transmitter. Club stations, even if operated by one operator during the contest, are in the multi-operator class. Multi-op/multi-transmitter entries are to use separate series of serial numbers for each band.

EXCHANGE:

Calendar

Sept 8*	DAFG Short Contest—VHF
Sept 8-9	North American Sprint
	ARRL VHF QSO Party
	WAE—Phone
Sept 9*	DAFG Short Contest—SW
Sept 15-16	CAN-AM Contest—CW
	Kentucky QSO Party
	Scandinavian Activity—CW
Sept 15-17	Washington State QSO Party
Sept 15 & 22	89'ers Run
Sept 22-23	Scandinavian Activity—Phone
Sept 29-30	Delta QSO Party
	CAN-AM—Phone
Sept 30-31	College Radio Scrimmage
Sept 30-Oct 1	Fall Classic Radio Exchange
Oct 6-7	QRP Annual October QSO Party
Oct 13-14	Arrowhead 50th Anniversary QSO Party
Oct 20-21	ARRL CD Party—CW
Oct 28	ARRL CD Party—Phone
Nov 3-4	Crazy Eight Net QSO Party
Nov 10-11	ARRL Sweepstakes—CW
	CQ-WE Contest
Nov 11	IPA Contest
Nov 17-18	OK DX Contest
Nov 24*	ARRL Sweepstakes—Phone
Nov 25*	DAFG Short Contest—SW
Dec 1-2	DAFG Short Contest—VHF
Dec 1-3	ARRL 160 Meter Contest
	Connecticut QSO Party
	North Carolina QSO Party
Dec 8-9	ARRL 10 Meter Contest

* = described in June issue

Continued on page 147



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I must first apologize somewhat for stealing this article's title from J. H. Everhart N2CX. His article, in the August, 1978, 73 Magazine, was entitled

"Power Line DX." But, I did not stop there! I stole even his circuits, almost intact. That, too, I must explain.

My intent is to praise him highly. The original ar-

ticle dealt with control of distant (DX) objects by remote control, via existing 60-Hz lines. He helped me by solving my worst problem in controlling my EME array, located on a tower building some 150 feet (by wire and zig-zag routing) from my basement ham shack. First, I built his circuits, and they worked — very well! Then I decided to lay out a PC board to speed matters up if I wanted more than one copy (I did). And then, I decided to do an article just for the PC boards, just to encourage others to try his methods — using his circuits — which, as I said, work very well indeed.

While laying out the first board and mulling over his idea for remote monitoring of his ham receiver, the way of laying out the PC boards shown in Figs. 1(a) and 2(a) evolved. Figs. 1(b) and 2(b) are just loading diagrams to hasten things along for you.

Fig. 1(a). PC layout, R/T board.

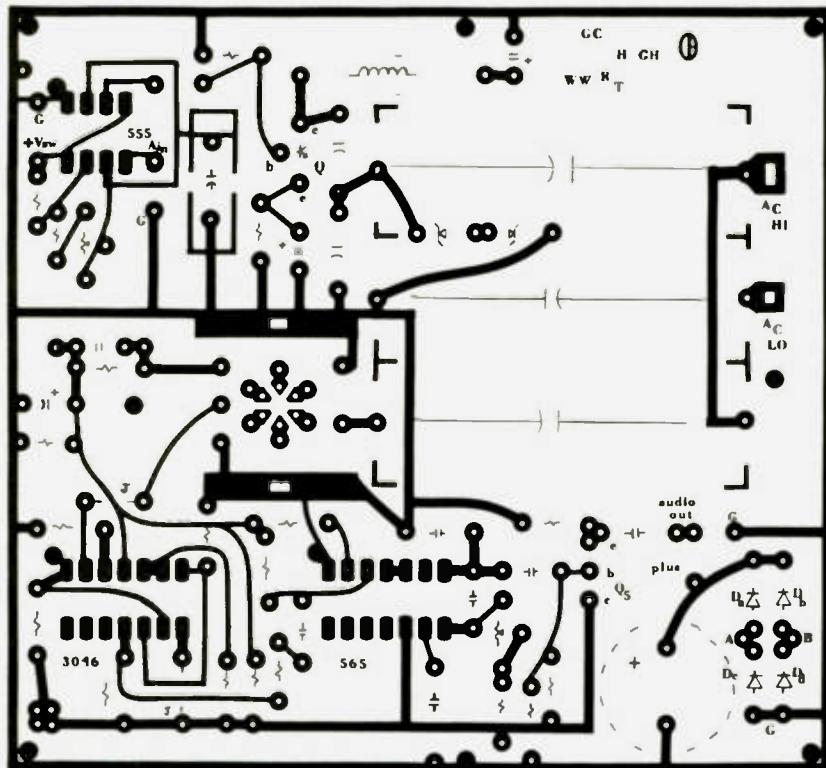


Fig. 1(a) is the PC board

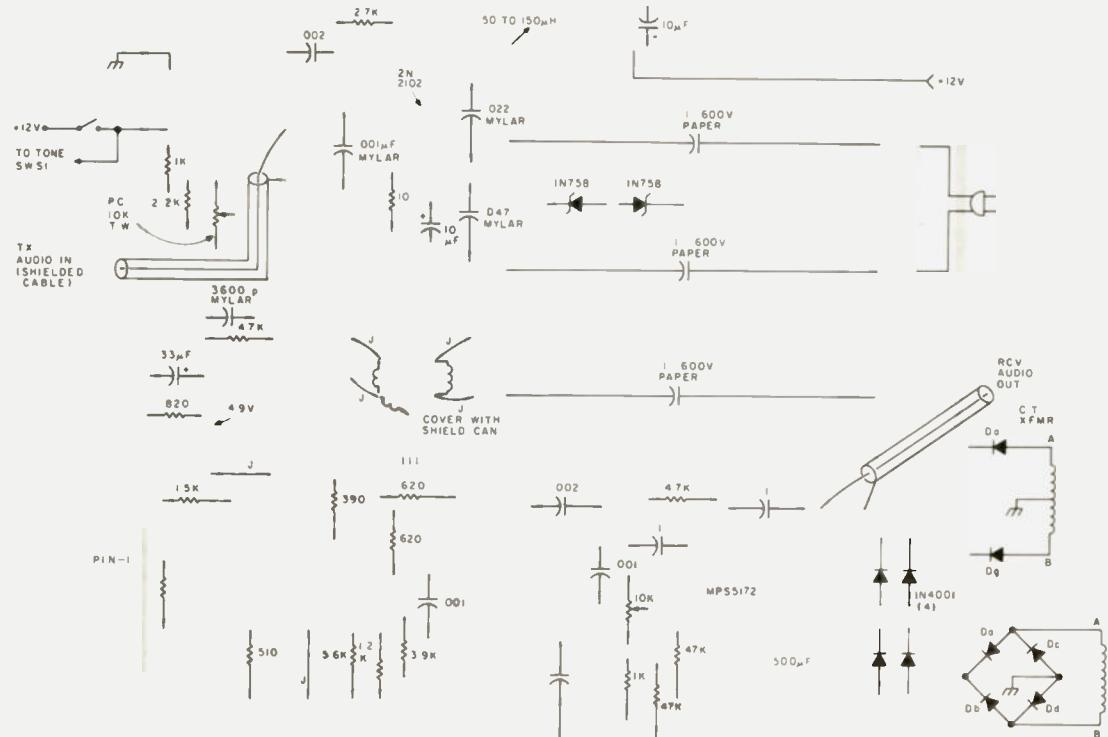


Fig. 1(b). Component layout, R/T board (foil side view).

layout for both transmitter and receiver, or what I call the R/T board. This is the heart of the system to allow hooking up to your power lines. Fig. 2(a) shows two boards—one that accepts up to three CD 4001

ICs and the components needed to handle the generation of up to six control tones (audio), and a second board that can be split up the middle. (See the vertical black line that appears to do nothing elec-

trically.)

The left half of this

layout is the audio amplifier (follower, really) and

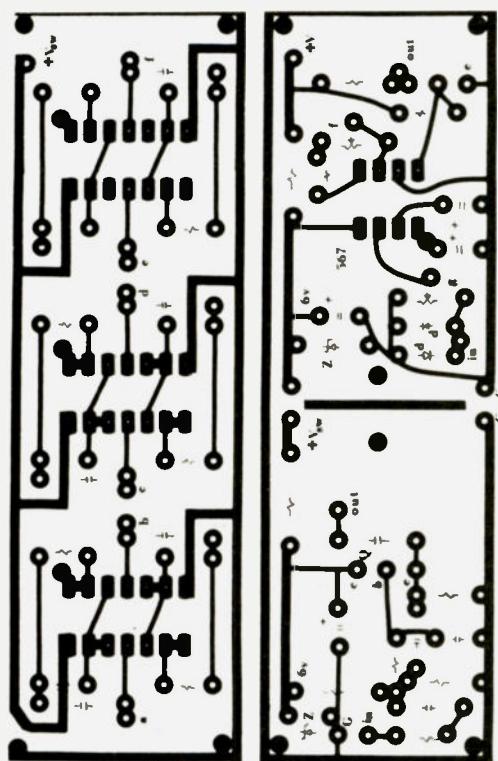
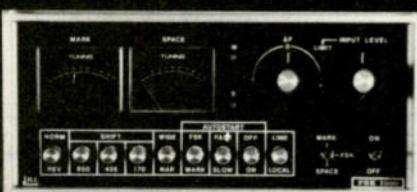


Fig. 2(a). PC layout, tone generator and audio amplifier/decoder-buffer boards.

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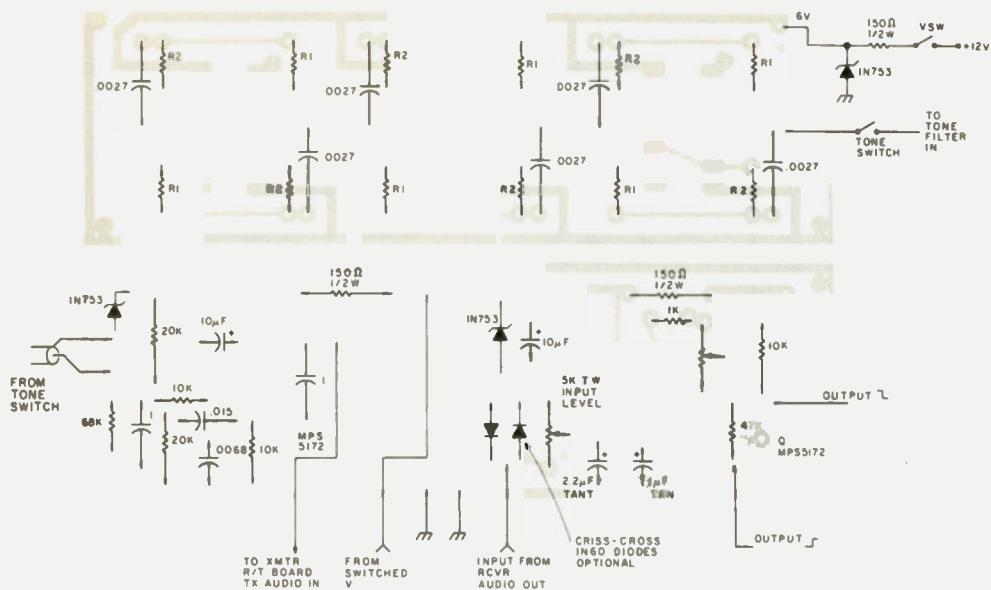


Fig. 2(b). Component layout, tone generator and audio amplifier/decoder-buffer boards (foil side view).

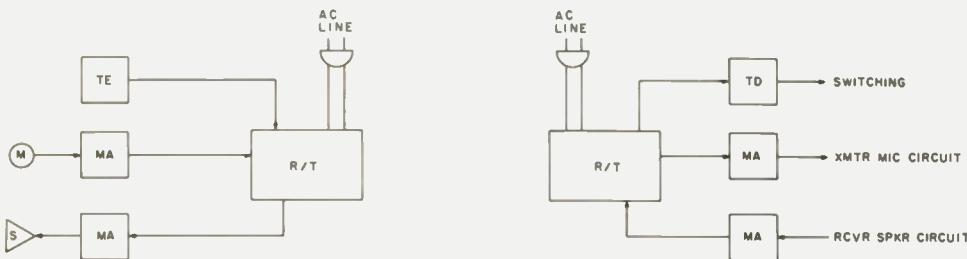


Fig. 3(a). M = microphone. S = speaker. MA = amplifier and matching transformer, if required. TE = tone encoder. TD = tone decoder. R/T = receiver/transmitter board (Fig. 1).

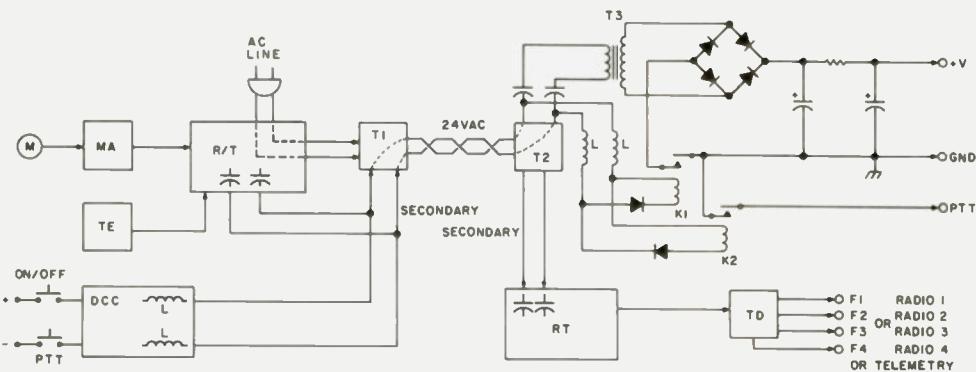


Fig. 3(b). DCC = dc control unit (as required).

X #	Tone	X #	Tone
L 1	69.7 kHz	H 1	120.9 kHz
L 2	77.0 kHz	H 2	133.6 kHz
L 3	85.2 kHz	H 3	147.7 kHz
L 4	94.1 kHz	H 4	163.3 kHz

Fig. 3(c).

the low-pass filter to follow the tone generators. It is shown as an MPS5172 in Everhart's article, Fig. 2, page 53. The right half of the board is all the circuitry to decode and/or buffer/invert a received tone. It

is the same type MPS5172 (Q6) and NE567 shown in his Fig. 5.

Once these Figs. 1 and 2 of my PC layout have been done in the photo negative stage, they may be cut apart and boards made in any combination or quantity. The PC board layout removes all the headaches and a lot of the work of this job, so I hope more of you will try N2CX's ideas.

I planned Fig. 1, the R/T

or outbuildings where there is a wall outlet.

Build up two complete R/T units like that and you have one wireless intercom that can be carried about from room to room and plugged in, while the second unit babysits your child or your ham gear for you. The R/T board contains all the necessary components to take audio in and give rf output ready for, and attached to, the power line in transmit—and the same in reverse, in receive. All that is needed in addition is some obvious switching and an audio amplifier. This would boost up a microphone level that is too low (1.4 V required at the NE555 input) and a received audio that may be lower than you want. Then get the speaker and microphone of your choice.

I got around the latter by using an older Motorola speaker-microphone in one unit, with a coiled cord like those used on the FM handie-talkies. Use the small transformers shown to match up impedances where and when required by what you use. My audio amplifier was a 1-to 2-Watt per channel, with one IC that you often see in a child's record player of the less expensive variety.

To give you an example of how I use all this—I sample a small amount of audio from the speaker output lines through a shielded cable and feed it over to the one R/T board I have built into my audio control unit in my console. The audio unit allows me to add tones, touch-tones™, and the like to my rigs, and also has decoding for the receive side of things. By adding the R/T board to the unit, I now can place any of these tones, voice from the station microphone, or audio from the receivers into the transmitter portion of the R/T board and ship it anywhere on my property.

where 120 V ac, 60 Hz is run.

Out on the remote end, where you have taken the R/T unit, this audio is taken back off the power line, decoded if required, and/or amplified up to speaker level. It has been a ham shack-to-garage and tower-house intercom, a control unit from shack to tower house for EME array control (a 150-foot unwired blessing), and even a two-way setup back to the woods almost half a mile away—where all I run out there is a low voltage, 24-V ac line to avoid electrical code (and cost) problems. See Fig. 3(b).

The two R/T units, in that case, are merely coupled into the 24-V ac secondary of the ham shack end and decoupled at the 24-V ac primary of the woods end. The woods-end transformer then transforms back up (to the 48 V ac I needed for a 30 V dc and allowing for line drop from the ham shack to the woods). Direct current control is sent down the same two wires for a one pair/three duties system.

My only concession in the R/T use was to utilize different rf frequencies each way, so that I can override receiver audio coming my way when I want to transmit either audio or a control tone. For monitoring the local repeater from anywhere in your house, it is great! A single tone is sent along with the microphone audio while you are transmitting. After the audio is demodulated from the rf carrier and leaves the R/T board at the ham shack end, it goes two ways. The audio goes through a parallel trap to reject the tone, to a one-stage audio amplifier (later found to be unnecessary), and, via a pot, to the microphone input circuit. The pot allows me not to have to change levels in

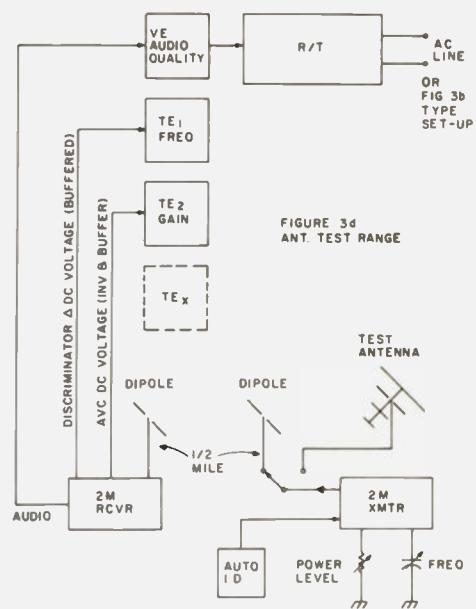
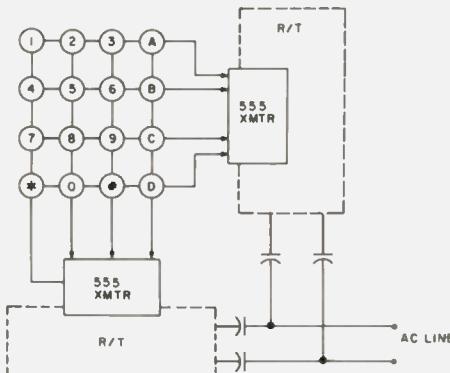


Fig. 3(d).

the radio or station microphone setup.

The second audio direction is for the tone; it is through a series pass filter, to reject audio and pass tone, and back to the one-tone filter and decoder you see in Fig. 2. That, in turn, when sensed and decoded, pulls in a relay and keys the transmitter. In the other direction, the audio mentioned earlier, from a receiver-speaker circuit, is coupled into the transmitter half of the ham-shack-end R/T board and put on the ac power lines. Back where I am (remote), the receiver portion of my R/T board demodulates the audio from the rf carrier, amplifies the audio, and presents it to me at the speaker.

I have added some complications since I want to monitor more than one radio and I don't just want to listen to noise at the remote end. On FM, that is not bad, as the radio has a squelch that keeps the audio off the transmitter input line of the ham shack R/T board, but I also monitor 50.125 MHz on SSB. To get around this, I have an audio-type squelch in the audio console that, when the radio speaks, keys a tone along with the audio

(sort of continuous-tone squelch style). When this tone gets to my unit (remote), it is handled just like the other direction, by splitting the audio and tone. The tone also lights an LED corresponding to what tone is controlled by which radio spoke. By using more than one tone at the remote end, I can control which radio I am putting the microphone audio into and keying. That about sums up one of my dandy uses, one of the reasons for writing this article, and the prime reason for my expanding it beyond just the PC board layouts.

I have changed some of the pin numbers of the ICs due to my board layout, so I have re-drawn the entire R/T board schematic as Fig. 4. The numbering on Fig. 4 and the layout now agree.

In the original article—in case they are not caught in the reader's column—let me say that there were a few errors and omissions. The pins on the "B" tone oscillator in Fig. 2 were omitted. I completely changed these, due to my layout, but in the original article, the left part should have inputs of 8 and 9 and an output of pin 10, and the right part, inputs of 12 and 13 and an output of 11.

Left for right numbering can be traded, if it makes your wiring easier, but don't accidentally trade inputs for outputs.

The four parts of the IC are identical in operation. In the original Fig. 5, the ICs were labeled incorrectly as 576s. They are the familiar 567s. And while I am on the 567s, beware of too much tone input. 50 mV should be maximum, if you want to preserve any narrowband characteristics. Q5, in Fig. 4, can supply much more than that. I have added limiter diodes to the circuits and a level pot that you can adjust on the PC board. If you have tone-falsing and -failing problems, act accordingly.

One other comment on the original article. Fig. 6 shows how to re-wind the rf transformer. While this worked well, it was a lot of work. I allowed, on the PC layout, for you to use a much more common coil form put out by many vendors through most supply houses. Wind the transformer on it in exactly the same manner as shown in the original article, cover with a matching small i-t coil shield, and the results will be the same. I also left the coil pinout the same, so don't forget the jumper

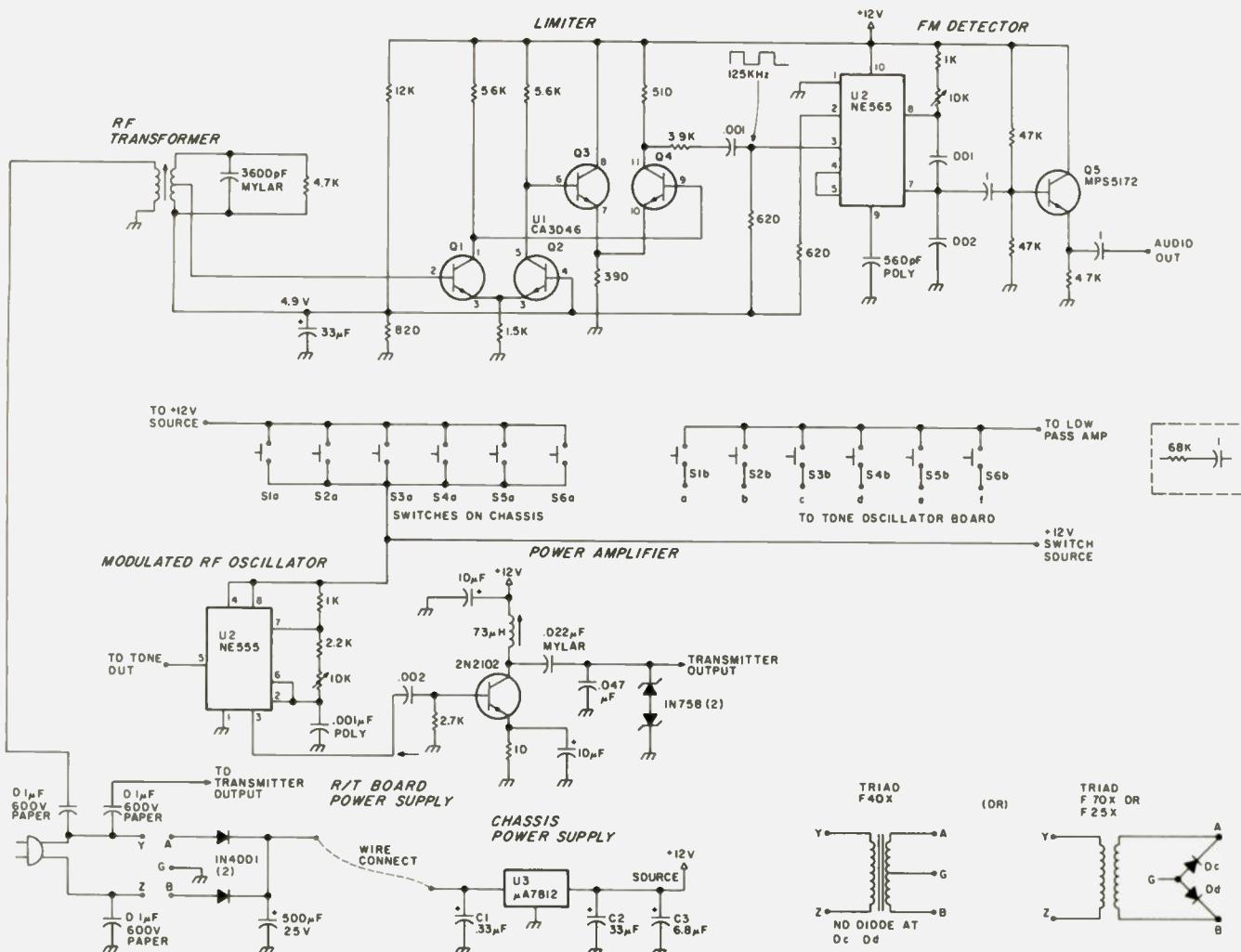


Fig. 4. Mount 7812 to chassis heat sink. Choose R_2 and C_X for desired tone. Recommended values are $R_2 = 100\text{k}$ pot and $C_X = .001$ to $.0047$. Trim pot around center and change C_X for coarse range.

wires (five are required) from the transformer base to each circuit. If any of the jumpers have to cross, be sure they are insulated wire.

I'll add in here a few useful notes on Fig. 3 concerning the hookups and the parts to use. Fig. 3(a) shows the simple form of wireless intercom unit used as a transceiver extension or remote. The mA units should be whatever matching transformers the microphone and speaker require to properly couple to the R/T board. The ham shack mA units are handled the same way, properly matching your rig with respect to microphone input and audio output (phones or speaker). Since this will vary so widely, I have chosen only to block-diagram it. Keep it as sim-

ple as possible at first, until you see what you really must have. For example, use a .1 capacitor from the speaker line to the R/T unit and from the radio audio gain-control to control-level. Add a pot for level later, and/or come off ahead of the rig gain control, to allow a quiet rig in the house with the XYL while you monitor from out in the garage or wherever you might be. Then add the audio amplifiers where required. There are many cases where audio amplifiers might not be needed at all.

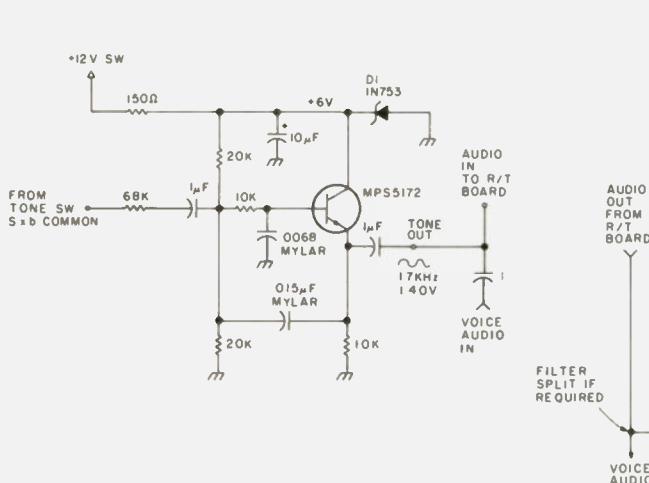
Fig. 3(b) is a special case.

I have found it very useful to reduce wiring between two points when there is not already a 120-V ac, 60-Hz line. See Fig. 3(b). If you run a safe wire, like rotor cable with good

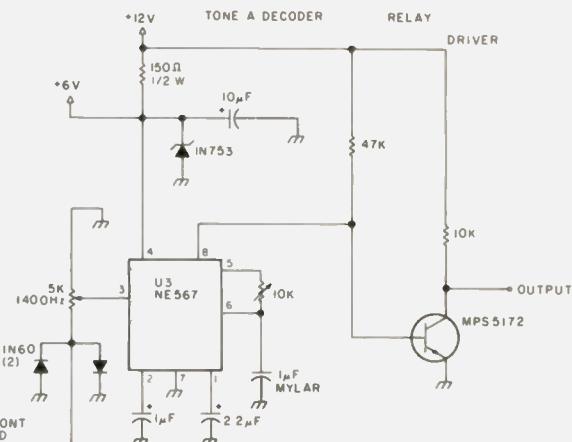
insulation, not many codes would prohibit such an installation. If 120-V ac wiring were run, all kinds of codes, regulations, electricians, and costs get into the act!

Skipping over Fig. 3(c) for a few paragraphs, since it is really a part of Fig. 3(d), Fig. 3(d) concerns a special situation not covered in Fig. 3(b). It allows a simple, remote transmitter-receiver pair to be located remotely from the ham shack/antenna location. The remote site then forms the heart of my antenna test range. Provisions are made for voice to return back over the wired circuit for both checking audio quality and intercom use when someone is at the remote site. Next, one channel (rf) of the wired remote is wired to use the buffered dc voltage from the receiver's discriminator (the VHF receiver). The buffer also shifts the level to allow direct use with the 555-modulation pin (all positive voltages), by shifting the discriminator minus 6 - 0 - plus 6 V dc (approx.) over to a 9- to 11-V dc shift. When the receiver is right on, the 10 volts then represent 0 V at the discriminator. Since the transmitter end (VHF) is where you are, the frequency is used to be sure you are sending right in the center of the receiver's (VHF) passband, where the receiver has been aligned, calibrated, and measured. That is, uV antenna strength equals some arbitrary avc reading, recorded. You use your reference dipole—see Fig. 3(d)—to

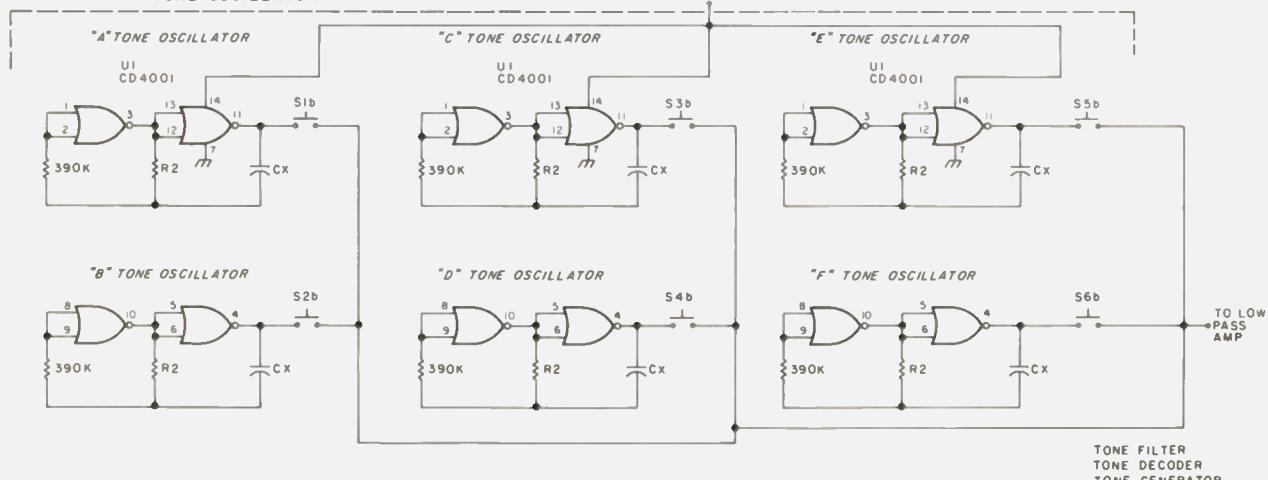
TONE LOW PASS AMP



TONE DECODER



TONE OSCILLATOR



the remote-site dipole first each time, and if the frequency reads all right, then any differences should be in the path (not likely at $\frac{1}{2}$ mile, but weather affects it some) or in the receiver itself, due to some factor such as temperature or aging. Then you proceed with your test antenna, compared to previous antennas, and the reference dipole using the avc voltage reading on a separate rf-wired channel.

Using the separate wired channels of rf allows continuous monitoring of all, simultaneously. I intend actually to monitor three voltages (plus frequency) in this measurement, along with a low-power transmitter whose output I can both accurately measure and vary continuously from 1 to 10 Watts. The remote receiver can be any FM unit on the frequency

of your antenna interest, but preferably solid state to reduce power consumption on your 24-V ac line out to the site. Then you measure and send back (by three rf channels) the two limiter and avc readings.

By varying the transmitter level, you can keep the receiver down out of limiting where the avc is linear or, at least, repeatable. Remember, you are not trying to recover audio from the VHF transmitter at the VHF site at this point, but only to compare the signal strength of two antennas: the reference dipole vs. your new model. The range is approximately $\frac{1}{2}$ mile, or 2640 feet, or 417 wavelengths on two meters, and gets better as you go up in frequency. This keeps reflections and stray ground clutter to a minimum, as

well as allowing you to plot your antenna, and any stray lobes, accurately. You require only a well-calibrated antenna rotor and readout. My remote site is due east (90 degrees) from me; that becomes 0 degrees on the antenna plot. The north rotor stop is a pain you live with, but I avoid extra work by starting at 270 degrees plot (due north—compass) and plotting around through straight at the site and through 360 degrees in 5-degree increments.

If you try this kind of work, always allow a full 360 degrees around—and then back in 5 degree increments. Trying two or three such trips and averaging the results is even better. This allows for any rotor over-travel or coast when stopping. Whatever you do when comparing antennas, or how you do it, never—

repeat never—believe the reports of signal strength changes from antenna to antenna that you get from another station. Modulation, keying, etc., is all right, but unless you like to climb towers, changing antennas back and forth and never seeming to get anywhere, stick with a measured system. This is not to say you can never have another amateur whose setup and serious attitude matches your own or the level of confidence you want, but to suggest you don't buy the casual-type observer. He may have changed rigs, antennas, and use his deaf ear for an S-meter! Why do you think I went to the trouble of a range?

Fig. 3(c) was thrown in to show how I can control 16 channels of rf using only eight frequencies and up to as many audio tones as you

can discriminate between or use without ending up in a crossfire mess. If the numbers and the control pad look familiar—they should! I claim no credit for this idea, beyond acknowledging that a lot of work and research (translation: money!) went into the selection of the touchtone™ frequencies used by the phone companies. Such things as harmonic relationships and

beats were all thought out for me; I just shifted the tones up to rf frequencies and used the same 567 decoder scheme used by many autopatch setups, but at rf frequencies. The 567 is good up to 500 kHz, so I am well within limits. Add to the fact that audio down a power line would compete with the 60 Hz and be hard to separate, and the 10% bandwidth at rf is several cycles you can

easily zero in on versus a few cycles at audio, and you realize why the shift up and the manner I did it.

Fig. 4, as stated earlier, is just schematics—the same, almost, as the originals electronically, but divided up as they appear on the PC boards and numbered in the PC board manner for easy troubleshooting.

I'm sure you will find at least as many uses for this

terrific circuit combination as I did, so I hope the PC boards make it a lot easier for you. Any other uses or ideas using these little gems I'd like to hear about, and an SASE is required only if you need a reply. Electronics or circuitry questions should go direct to N2CX, and PC board or uses questions to me, to keep things sorted out. Give remote control a try—you'll like it, I bet! ■

New Products

from page 24

tunity to save time and money in QSLing as well as have an instant reference right at his fingertips. Duplicate QSLing will become a thing of the past. You know at a glance what you have worked, on what bands and modes, and, most important, what needs to be confirmed. Each call area of the United States is divided by state, band, and mode. An extra column is provided for new

modes or notes. As an added feature, a section is provided for every country of the world. "Insta-Pac" is attractively packaged and would be an asset to any ham shack. It has been developed by KH6ITY and is offered with an unconditional money back guarantee if not completely satisfied.

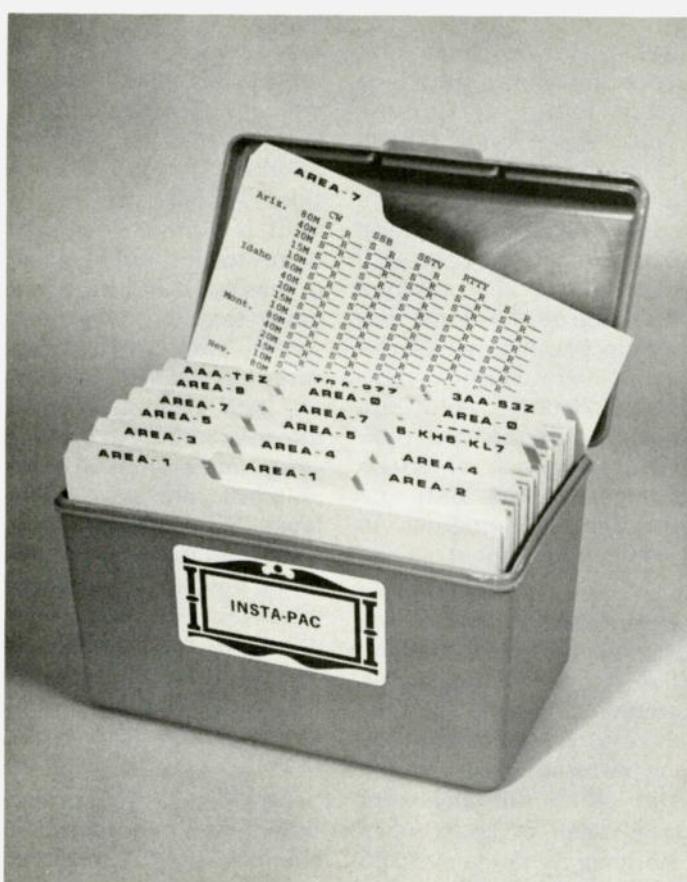
For further information about this new system, contact Insta-Pac, PO Box 22974, Honolulu HI 96822. Reader Service number 143.

NEW BEARCAT® POCKET SCANNER RECEIVES BOTH AIRCRAFT AND GROUND-CONTROL COMMUNICATIONS

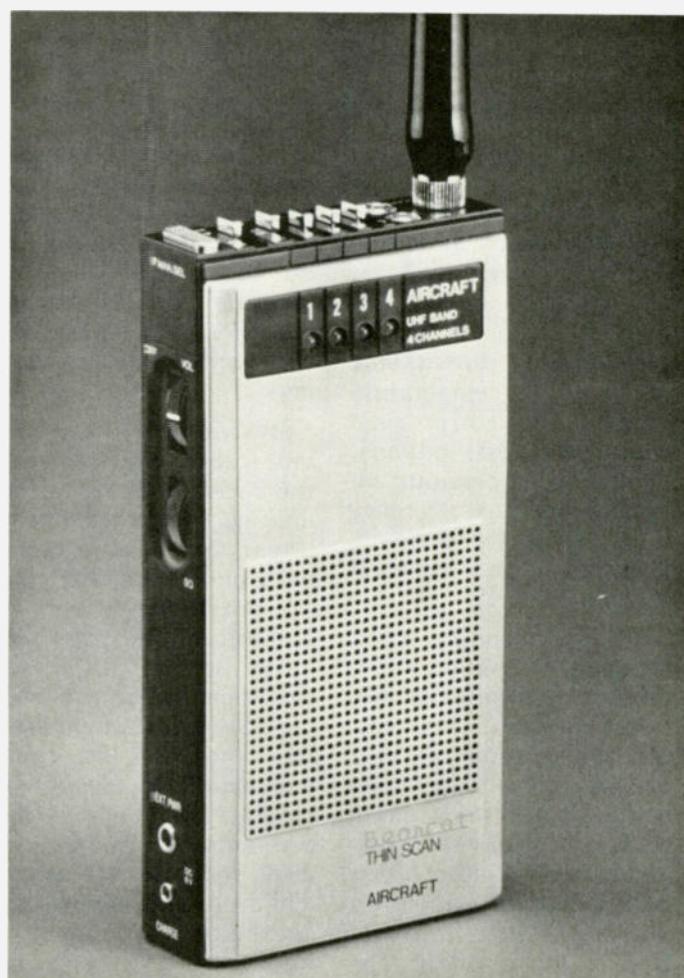
The world's first pocket-size scanner radio capable of receiving both the VHF-AM aircraft channels and UHF-FM channels has been introduced by Electra Company. For the first time, airport officials, operating personnel, pilots, and others interested in airport operations can carry in their pocket a single radio which can keep

them informed of both tower-to-plane and ground-control communications. The new "Bearcat® Thin Scan Aircraft" scanner is just 2 1/4 inches wide and 1 inch thick, and weighs a mere 10 ounces, making it a true pocket portable. The radio is capable of receiving up to 4 channels in any mix of the two bands covered. Each channel is provided with a lockout control for bypassing when desired.

Continued on page 157



The Insta-Pac.



The Bearcat® Thin Scan Aircraft scanner.

The FIRST choice by amateurs the world over!

Vhf engineering REPEATERS

4 MODELS TO CHOOSE FROM:

RPT 50	50 - 54 MHz
RPT 144	144 - 148 MHz
RPT 220	220 - 225 MHz
RPT 432	420 - 450 MHz



AVAILABLE ONLY AT THESE AUTHORIZED DEALERS

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Vhf engineering
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Blueprint for Biofeedback Experimentation

— fertile ground for pioneers

The idea for this article came from an experiment into the capabilities of human movement and control. The main question was how it would be possible for a person to control electrical devices and effectively communicate without movement of any kind. This restriction included the use of speech, sound, or any alteration of respiration.

Upon first consideration it seemed an impossible task, as the only apparent control would be over thought. This is where the idea of control by brain-wave came to mind. No, I'm not talking about ESP or psychokinesis, but biofeedback. Biofeedback is the process of monitoring and controlling certain body functions...in this case, alpha brain waves.

The main purpose of biofeedback is to provide greater relaxation and

creativity by controlling certain types of brain waves. In this article, we will be concerned with alpha waves.

A device called a "biofeedback monitor" is used to amplify and detect alpha waves in much the same way as an electroencephalograph. A band containing electrodes is fitted around the head, and when the brain is producing an abundance of alpha waves, the monitor gives indication with a tone or light. By learning to control the tone or light, you learn to control your alpha waves.

At this point, I'm sure you are beginning to see the *modus operandi* of the alpha control. Fig. 1 shows a simple block diagram demonstrating the basic principles involved. In this diagram, a photocell light and its counterpart, a tone-voice-operated microphone,

phone, are used to show a working system in its simplest form.

While the system described would work, it was not made to replace electronic switching and logic circuits, which would be more accurate. Schematics for an actual control circuit have been omitted, as the design and construction would vary greatly with the type of monitor output.

There are three major types of outputs used in the majority of monitors. They are: threshold tone, frequency-modulated tone, and light or LED indicator.

As for the monitor itself, you can get it pre-assembled or in kit form. There are models with simple one-mode outputs and ones with more complex multi-mode outputs. The range of control can be increased by the addition of a transmitter connected to the

monitor and a receiver connected to the actual control circuit. A wireless microphone and receiver would work nicely, and possibly allow the utilization of existing equipment.

The next logical step would seem to be to develop a means of controlling more than one device. One possible but untested idea would be the use of a row of lights that would flash in sequence, somewhat like a public band scanner. By having the lights represent controlled devices, or letters, any number of appliances might be controlled simply by stopping the scan on the desired light and device it represents. Of course, the speed in which one could control his alpha waves would be a factor in the speed of the scan. In such a manner, a person totally paralyzed could communicate and, to some extent, control his environment.

It has been my honest effort to present these ideas in more than the light of a laboratory or scientific curiosity, and I now leave further experimentation and development to your able ingenuity. ■

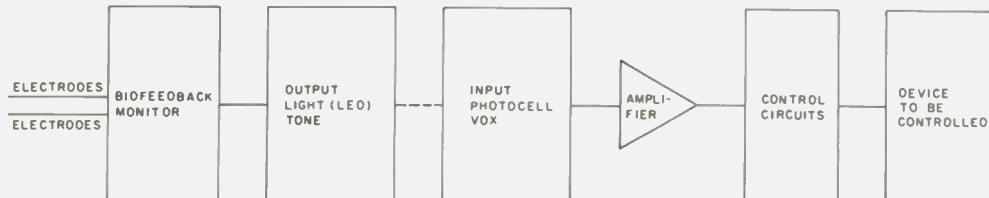


Fig. 1.

The Scanning Memorizers



FT-127RA

(220 MHz)

FT-227RB

(144 MHz)

FT-627RA

(50 MHz)

The FT-127RA, FT-227RB and FT-627RA, FM transceivers, allow scanning and expanded memory coverage for the demanding VHF FM operator. All feature up/down scanning capability with control from the microphone; the scanner will also search for a busy or clear channel. Four memory channels are available — two for simplex, three for repeater channels, one for a split of up to 4 MHz. Other performance features are similar to those of the renowned FT-227R.

OPTIONAL EQUIPMENT

Keyboard Microphone: YM-22 for FT-127RA and FT-627RA

(YM-22 standard feature with FT-227RB)

Squelch Unit • FP-4 AC Power Supply

CPU-2500R/K 2 M FM Transceiver with Central Processing Unit

The age of computers has entered the amateur scene with the announcement of the CPU-2500R/K 2-meter FM transceiver. Controlled by a 4-bit central processing unit (CPU), the CPU-2500R/K contains a scanner, 4 memory channels, manual or automatic tone burst, an optional sub-audible tone squelch, and 25 watts output.

The keyboard microphone allows two-tone input for autopatch or control purposes, as well as remote programming of dial or memory frequencies.

Automatic ±600 kHz repeater split, or program a split up to 4 MHz using the memory. Keyboard microphone allows remote programming of odd splits.

CPU scanner will search for a busy or clear channel, upon your command.

Four memory channels for simplex or repeater use, plus another memory channel for a split of up to 4 MHz.



879

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The History of Ham Radio

— part X

Reprinted from QCC News, a publication of the Chicago Area Chapter of the QCWA.

The Second Radio Telephone Conference, called by Secretary Hoover, convened on March 20, 1923, to solve the deplorable situation brought about by the controversy over wavelength assignments. Broadcasters, the amateurs, the Army and Navy, the ship operators, and the commercial interests could not come to a common agreement as to wavelength allocations and thereby avoid woeful interference. Clearly, from an administrative standpoint alone, the broadcast listeners were desperate. The impasse which existed required a solution.

After several months of meetings, serious discussions, and lengthy deliberations, not even the amateurs had guidance or a clear go-ahead on how to interpret their standing with the Commerce Department. The American Radio Relay League had made a series of suggestions. The amateurs in all parts of the

country, now almost 21,000 in number and licensed to operate in the narrow 200-meter range, found solace in the fact that they had pooled their ideas and suggestions through their re-

spective district directors.

The ARRL Board of Directors based high hopes on all of these inputs. So, when recommendations were issued on June 28, 1923, by the Department of

Commerce in the form of General Letter #252, directed to all government district radio inspectors, they read as follows:

General Letter #252

"General and Restricted Amateur Radio Station Licenses will be issued permitting the use of any type of transmitter (CW, spark, ACCW, ICW, SC, and phone) with the restriction that when using pure CW, they are authorized to use wavelengths from 150 to 200 meters. When using spark, ACCW, ICW, unfiltered SC, and phone, the wavelengths from 175 to 200 meters only can be used. The types of transmitters must be specified in the application and the license.

"Special Amateur Radio Station Licenses will be issued permitting the use of pure continuous wave transmitters only, authorizing the use of wavelengths from 150 to 220 meters.

"For the purpose of application of Amateur Stations, pure CW is defined as follows: A system of telegraphing by continuous oscillations in which the power supply is substantially direct current as obtained

C.W. Licenses

Do you fellows know that your station license provides that the apparatus described in the application shall not be changed without permission? And that a license granted a spark station is not good for the use of C.W. equipment? It sounds queer, considering that any complainant ought to prefer a tube set to a spark, but if anybody wants to "get" you on it they can.

A broadcast listener recently filed a complaint with a radio inspector against a well known eastern spark amateur, alleging the use of an illegally broad and too-long wave. Called upon the carpet, there was no evidence that his spark, which was good as such animals go, was not entirely legal. But he volunteered the information that he was experimenting with C.W. anyway and the interference probably would be still further reduced. Whereupon, altho his spark set was O.K., he was informed that the use of C.W. was in violation of his license, and it was suspended for three months.

Moral: If the equipment you are using is greatly different from that for which your license was issued, take up the matter with your Inspector and "get right."

from (1) a generator, (2) a battery, or (3) a rectifier with an adequate filter. (A filter is not deemed adequate if the supply modulation exceeds five percent.)

"On licenses issued for Amateur Stations you will include the following: 'This station is not licensed to transmit between the hours of 8:00 and 10:30 p.m., local standard time, nor Sunday morning during local church service.'

"Special Amateur Stations must be operated by persons holding an extra first class grade Amateur operator's license, or a commercial extra first class operator's license. Applicants must also meet the requirements of Regulation 63.

"A new class of Amateur operator's license is hereby established, to be known as 'Amateur Extra First Grade.' Licenses of this grade will be issued to persons passing the required special examination with a percentage of at least seventy-five and code speed in sending and receiving at least twenty words per minute, five characters to the word; who have had at least two years experience as a licensed radio operator; and have not been penalized for violation of the radio laws subsequent to the date of these regulations."

The government supervisors, located in the several United States radio districts, upon receiving the new regulations, were requested to notify all amateur radio licensees, general and restricted, to submit their respective license papers to the supervisor's office and have them modified in accordance with the new regulations. After they were so certified with a copy of the new rules appended, they were returned, officially endorsed, extending the wavelengths range and specifying the quiet hours. No changes

were made in their expiration date.

All amateurs who held special licenses were notified that their licenses were cancelled and new licenses would be issued in accordance with the new permits now granted. Up to the present, all amateur licenses were issued specifying the wavelength of the operating transmitter to be used. From this date on, a license permitted a station to change the operative wavelength to accord with the bandwidth permitted and as the type of emission specified.

Variable frequency transmitter circuits were not generally known among amateurs. The progress of the art and the know-how was still lacking in 1923 in this respect. Even the well-informed and up-to-date amateur solved the changing frequency problem by having more than one transmitter available, usually a pure CW of low power and another with either ICW or ACCW plate supply. Every license issued by the Commerce Department required that the licenses specify not only the wavelength limits of the transmitter, but also the emission type and the apparatus to be used including the antenna length and construction. If an amateur qualified as an extra grade operator, he could apply for two licenses, a special and a general. The special was then given a "Z" call.

The ARRL had recommended that the amateur with two years of experience and a twenty-words-per-minute code speed be issued an extra first grade amateur operator's license using a wave transmission length up to 220 meters. (Even at this date all authorities assumed that this stipulation was a decided advantage for DX.)

It also was understood that transmissions on wave-

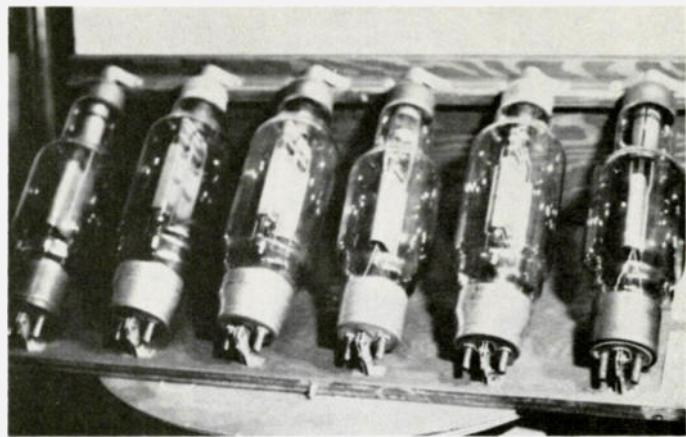


Fig. 1. The 250-Watt UV204.

lengths other than 200 meters by the amateurs could be allowed on the issuance of a license specifically noted on the application. A second wavelength, perhaps somewhere between 175 to 180 meters, could be granted. The 150 to 220 meter band specified in the 1923 Regulations gave the Secretary authorization to grant licenses upon request provided the amateur could meet the stipulated requirements and so requested in his application.

How the Need for "Plate" Power Was Met

In the early wireless days, the conversion from simple spark coil to high voltage rotary-gap transmission was not difficult. But going over to the vacuum tube for CW operation was by no means an overnight accomplishment. To put continuous wave power into the antenna, the amateur initially had available a small amplifier tube and a larger five-Watter, known as the UV202. Subsequently the fifty-Watt 203 was available, followed soon by the 250-Watt UV204, a real "power-house" for most DXers (Fig. 1).

With the introduction and development of the power tubes, there naturally was need for a direct current supply to energize the plate circuit. Much of the spark gear the amateur had accumulated and used for

a power source was now destined for the surplus stockpile. The new regulations specified pure CW or nearly so, and this necessitated some type of direct current source, often up to several thousand volts. Various methods to obtain this voltage were outlined in the ham literature: (1) a motor generator, or (2) a battery, or (3) a type of rectifier with filter. Much experimenting took place. This voltage was referred to as the "B" voltage supply.

The Motor Generator

Although not the most practical source of high voltage dc nor the least expensive, a generator was the easiest and simplest way to quickly come up with plate power of one- to two-thousand volts. The ESCO machine from Electric Supply Co., of Stamford CT, could be found in many ham shacks. Early commercial broadcasters also used this equipment. However, the supply required filtering to reduce the ripple modulation to a point where it would meet the specified five percent. Using a motor generator required ac starting controls, considerable power wiring, and usually a remote and out-of-the-way installation of the entire unit to reduce hum and noise when operating voice modulation. Weight and expense were



Fig. 2. A one-kilowatt transmitter.

responsible for the few installations of this kind.

Edison and Lead-Acid Batteries

Edison-type batteries were available built in small but compact units.

They were kept charged with a Tungar charger-rectifier connected to house current. The lead cells could be obtained in compact assemblies giving sufficient wattage for the smaller transmitting tubes.

All these units were messy and, besides giving off a gas odor, ruined many floors and carpets.

Typical of what certain amateurs were capable of and determined to build is illustrated by the enviable installation of radio NU9BHT. This beautiful layout required assembling and building one thousand individual lead-acid cells. They were contained in $4\frac{1}{2}'' \times 5\frac{1}{2}''$ hard rubber cases, with one positive and two negative plates per cell. Hard rubber trays contained 25 cells each and were treated with acid-resistant paint to provide protection and longevity. With each bank delivering 50 volts, separate switches were used to make appropriate connections for charging each bank, either separately or in four bank series of 200 volts. A motor generator battery charger delivering 250 volts kept the batteries fully charged.

Installations of this type necessitated adequate ventilation, care, considerable

service, and a separate building to house the equipment. The circuit diagram of the station transmitter adapted itself to frequent modifications and changes to keep it up-to-date.

With the development of crystal control and other experimental features added from time to time, NU9BHT established an enviable record while in service. Operating primarily on 40 meters CW and occasionally on 20, as an official ARRL relay station, it contacted all Australian districts, making five-continent QSOs in one evening. This was an exceptional accomplishment back in 1923. Not many installations of this type prevailed in hamdom. With a battery voltage source, no filtering was required. This was pure dc!

The Rectifiers, Mostly Chemical

Over the next few years, the various amateur publications carried a series of suggestions for methods of building rectifying equipment for the B supply.

How did the amateur go about putting together a chemical rectifier? He was on the lookout for—item one: chemically pure aluminum sheets; item two: chemically pure sheet lead; item three: a quantity of pint-sized jars; item four: several pounds of borax; item five: plenty of distilled water.

For each cell assembly he figured that approximately 40 volts could be rectified from a 60 cycle source. This translated to about 50 cells required for a 2000-volt output. Forming and placing such a home-built rectifier initially into operation required a great deal of experimenting to overcome a whole series of mishaps. Rectified ac output of this type was always accompanied by a 60-cycle ripple hum, which

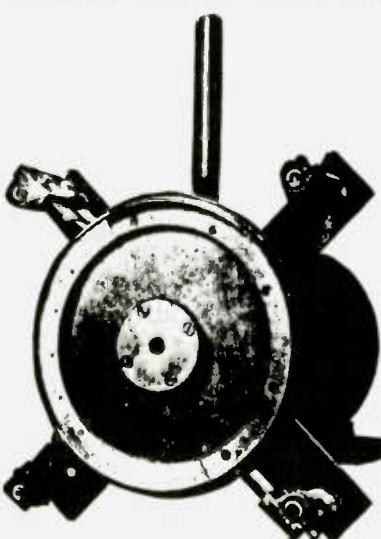
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It is the only rectifier that is adaptable to either high or low power sets as it easily handles up to 4,000 V. at 250 M.A.

The commutator on the Super is eight inches in diameter and is driven by a $\frac{1}{4}$ H.P. 110 V. 60 cycle 1800



RPM synchronous motor.

With the Super there are no materials to change—just connect the motor and high voltage leads and forget about it as the only attention required is an occasional oiling of the bearing.

The Super practically eliminates interference caused by other types of synchronous rectifiers.

Write for descriptive literature.

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Fig. 3. A synchronous rectifier.

With a half inch spark coil a hard receiving tube may be used; either the Cunningham C.301 or the Radio Corp U.V.201 is suitable. For larger coils the 5-watt C.302 or U.V.202 may be used but less filament power is consumed by the Western Electric "E" or VT-1 tube. No 50-watt tube with proper plate insulation for spark-tube work is available just now but the C.303 or U.V.203 can be made to do.

Various circuits have been described in our pages in a paper called "A Spark Coil C.W. set," by Samuel Kopelson of 2ECF (May 1922—p. 66), and in "A Spark Coil C.W. Transmitter," by F. J. L. Duffy (March 1922—p. 28). In *The Modulator* for January 1923 there appeared on page 17 a concise paper by M. Joffe describing his spark-tube set at 2RYO.

The best plan of all is to remove the secondary entirely and re-wind the coil with wire 6 or 8 sizes larger. Robert Kraus of 2CEI finds that for a Ford spark coil the best secondary is one having 5000 turns of number 32 A.W.G. (B&S) double-cotton-covered wire. When the secondary is re-wound in this fashion no shunt condenser is needed in Fig. 4 and even for the other circuits it can be reduced to .001 microfarad. Three sheets of 4 x 5 tinfoil between 5 x 7-inch glass sheets $\frac{1}{8}$ inch thick is more than enough.

had to be adequately filtered before it could meet the prescribed specifications—especially if it was to be used for a near-pure CW purpose. Filters were needed to eliminate key clicks. With batteries, there was complete absence of blinking lights, buzzing, whining, vibrating generators, and many ac growls on the air.

Other Substitutes For B Supplies

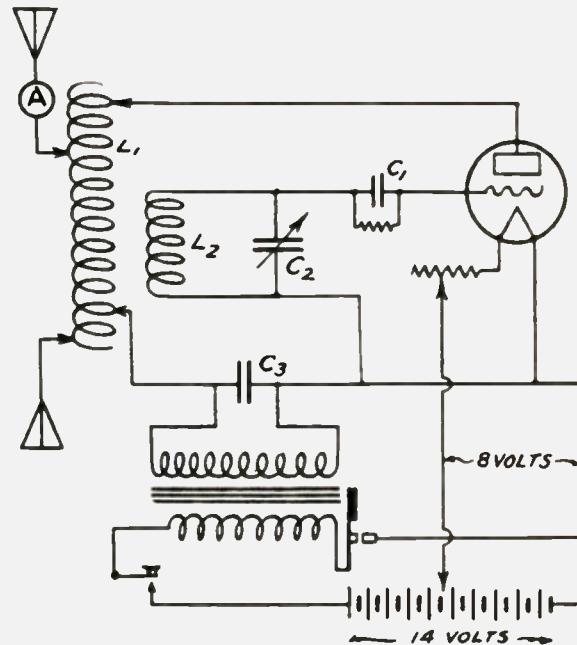
A convenient and inexpensive unit known as the mercury arc rectifier, equipped with an electrolytic "keep-alive" mechanism, was extensively used to give up to 3000 volts of B supply power. Both voltage and current outputs satisfied a one-kilowatt transmitter (Fig. 2).

Also available was a synchronous rectifier, as shown in Fig. 3. A synchronous motor, 1800 rpm, carried a large-diameter split

wheel, with which two husky commutator brushes made contact. In this way, the 60-cycle current was rectified and the pulsating directional output then filtered. These units were marketed under the name of Super-Sync and provided plate energy rated at up to 4,000 volts.

Some of the early plate supplies were obtained from specially-wound spark coils—see Fig. 4. Such transmitter assemblies put out an ICW signal, were extensively used, and solved the high-cost problem. It was necessary to adjust the vibrator frequently to obtain a smooth tone, the pitch being immaterial but often adjusted to satisfy the contacted operator at the other end.

Radio amateurs in the early 1920s were an experimental and ingenious lot. They solved their problems in the best amateur tradition. ■



L1—25 turns bare No. 10 wire on a 5 1/2 inch tube.
 L2—20 turns No. 18 S.C.C. copper wire on a 3 inch tube.
 C1—Mica condenser of .00025 to .0005 microfarads capacity.
 C2—Variable condenser, .0005 microfarads.
 C3—Glass or other sending condenser.

Fig. 4. Specially-wound spark coils.



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Maximum Security for the 22S

— remote control puts your Icom in the trunk

Probably one of the most versatile and easily modified radios on the market today is the Icom 22S. A multitude of frequency combinations can be selected simply by soldering inexpensive diodes into one of the 22 matrix board divisions.

The purpose of this article is to look at a few easy and inexpensive modifications which can be done to allow either greater frequency selection flexibility or remote control of the Icom "Mobile Marvel."

Upon examining the Icom 22S, one of the first things one usually notices is that, although the radio is advertised as a 22-channel rig, the programming matrix

board and selector switch have 23 positions. To initiate 23-channel operations, all one need do is connect a short piece of wire from switch lug 23 to the 23rd diode row and program the row with appropriate diodes for the extra frequency of your choice.

The second easy modification is the use of the 23rd position in conjunction with a few diodes and a DIP switch to allow the selection of any combination of available frequencies. Articles in the 73 Magazine issues of June, 1977 (p. 152), and May, 1978 (p. 158), showed DIP switch mounting methods which required external components subject to easy

breakage, or required defacing the cabinetry of the radio. An alternate method which allows easy DIP switch programming without defacing the exterior of the radio, while providing a secure and strong DIP switch mount, is shown in Fig. 1.

All one needs to do is simply (1) remove the 24-pin accessory plug from its holder, cover it with tape to prevent shorts, and push it inside the case, and (2) glue a section of perf-board on the accessory plug support bracket, mount your DIP switch on the perfboard, and wire as shown in Fig. 2. At this point, your 22-channel radio can be programmed for any of the frequencies

listed in the owner's manual, without resoldering diodes.

The final modification I will describe was inspired as much by my laziness as by security needs. Often I find myself needing to stop in town for some small item. As in most urban areas, this requires that either you remove your radio from the car or trust fate that it will be there when you return.

Not being one who enjoys disconnecting my rig and placing it in the trunk four or five times a day, I conceived the idea of remote controlling the Icom and mounting it in the trunk.

To my great surprise, this modification is both easy

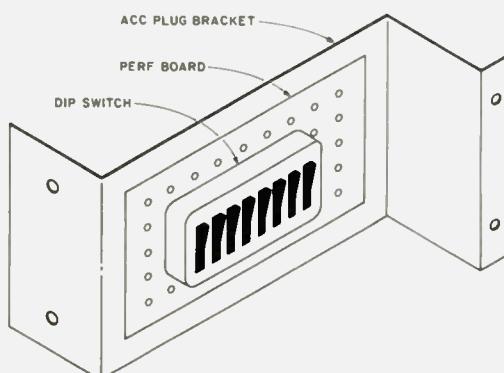


Fig. 1.

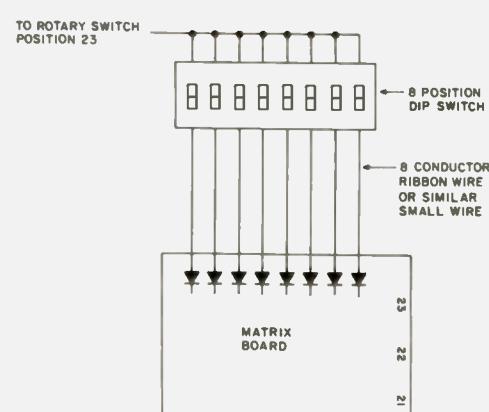


Fig. 2.

and inexpensive. All that is required is some wire, two potentiometers, two knobs, one male and female 4-pin mike jack, a speaker and miniature phone jack, a rotary switch, and a small box, such as the Radio Shack #270-236.

To accomplish remote frequency selection, run a wire from an unused channel position on the matrix board (Point A in Fig. 3) to the common of the remote 23-channel selector switch, via the accessory jack, as shown in Fig. 3. Note that the previously-mentioned DIP switch location cannot be used with remote control. It will be necessary to mount your DIP switch elsewhere (see Leon Baldwin's article in the May, 1978, issue of *73 Magazine*). Next, run your return leads from the rotary switch back to the matrix board via the accessory jack. All 22 previously-programmed channels cannot be remotely selected due to the use of the accessory jack to carry volume, squelch, and duplex controls.

If you do not have a 23-position rotary switch, an alternate method may be to use a double-ganged, 12-position rotary switch and toggle switch combination, as shown in Fig. 4.

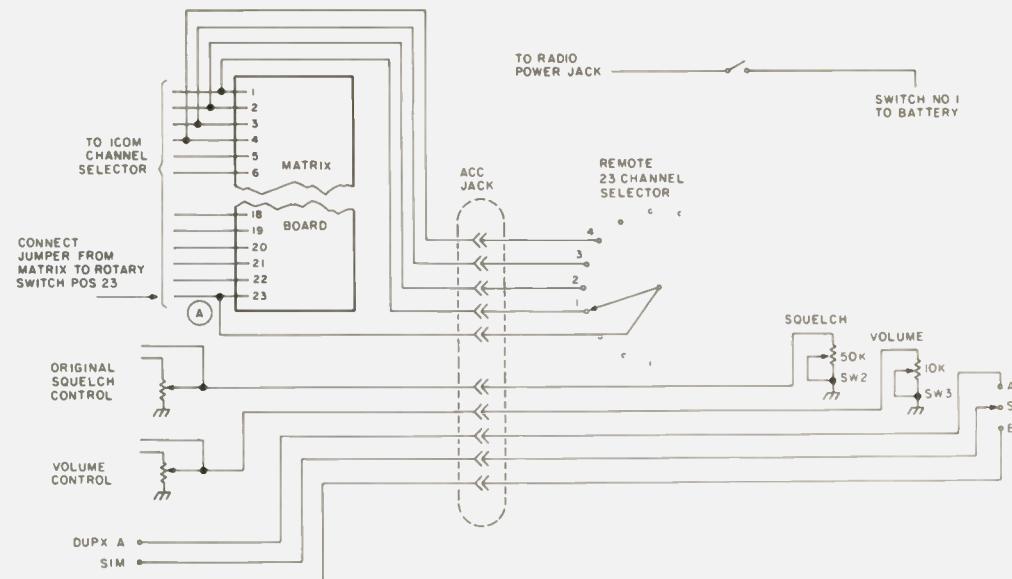


Fig. 3.

To vary the volume and squelch level, simply install a pot in parallel with the wiper on the original controls, as shown in Fig. 3. Slight problems may arise in the volume control circuit. Usually, this is due to a pot which will not adjust to a low enough value. It may be necessary to substitute a pot other than the one specified.

Duplex-simplex control also is accomplished as shown in Fig. 3. Power for the unit is delivered from switch 1 in the control panel.

To operate remote, install the unit with the power switch on the radio

in either high or low position. Turn the squelch controls on the radio to maximum, and the volume to about mid-point. Place the channel selector in position 23, and control as usual from the remote unit.

I hope that this article will stimulate interest in Icom owners to build the remote unit, and possibly to try other remote control modifications to which this versatile little unit might well be suited. ■

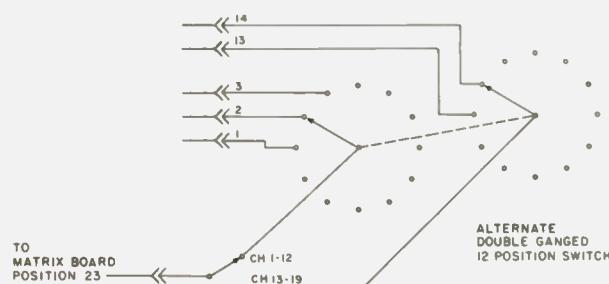


Fig. 4.

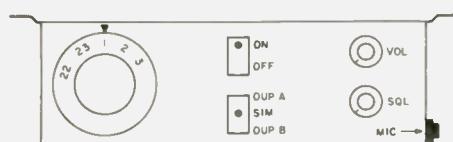


Fig. 5.

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The Big Blinker

— a visual signal for the deaf

A friend recently asked me for something which could be used in conjunction with an alarm clock in order to awaken his deaf nephew in the morning. The circuit which I finally came up with is shown in Fig. 1.

Basically, it works as follows: The sound of the alarm is picked up by a crystal microphone element, amplified by U1A and U1B, and detected by the full-wave rectifier consisting of U1C and diodes D1 and D2. The rectified

voltage is next applied to a comparator (U1D). When the input signal is loud enough and of a long enough duration, the rectified voltage on the inverting input of U1D exceeds the reference on the non-inverting input, causing the output of U1D to go to Vee. This turns on transistor Q1, which then enables U2, a 555 timer operating as a one-second oscillator. The output of the 555 drives a triac, which in turn causes a lamp plugged into the ac socket to flash

at a one-second rate.

For best results, tape the microphone to the alarm clock and adjust the comparator threshold so that only the alarm clock sound triggers the unit. Transient noises will not enable the circuit due to the RC time-constant at the detector output.

All of the parts are readily available at your local Radio Shack. I used an LM324 quad op amp, but any quad, two dual, or four single op amps can be

used. The pin numbers shown were for my particular layout but, of course, any of the op amps can be interchanged.

This unit performs quite well, and is certainly more of an attention-getter than the original timer used. The original timer was used to simply turn on a light. The triac used can handle up to 6 Amps, so several lamps can be plugged into the unit. The lamps will continue to flash as long as the alarm is sounding. ■

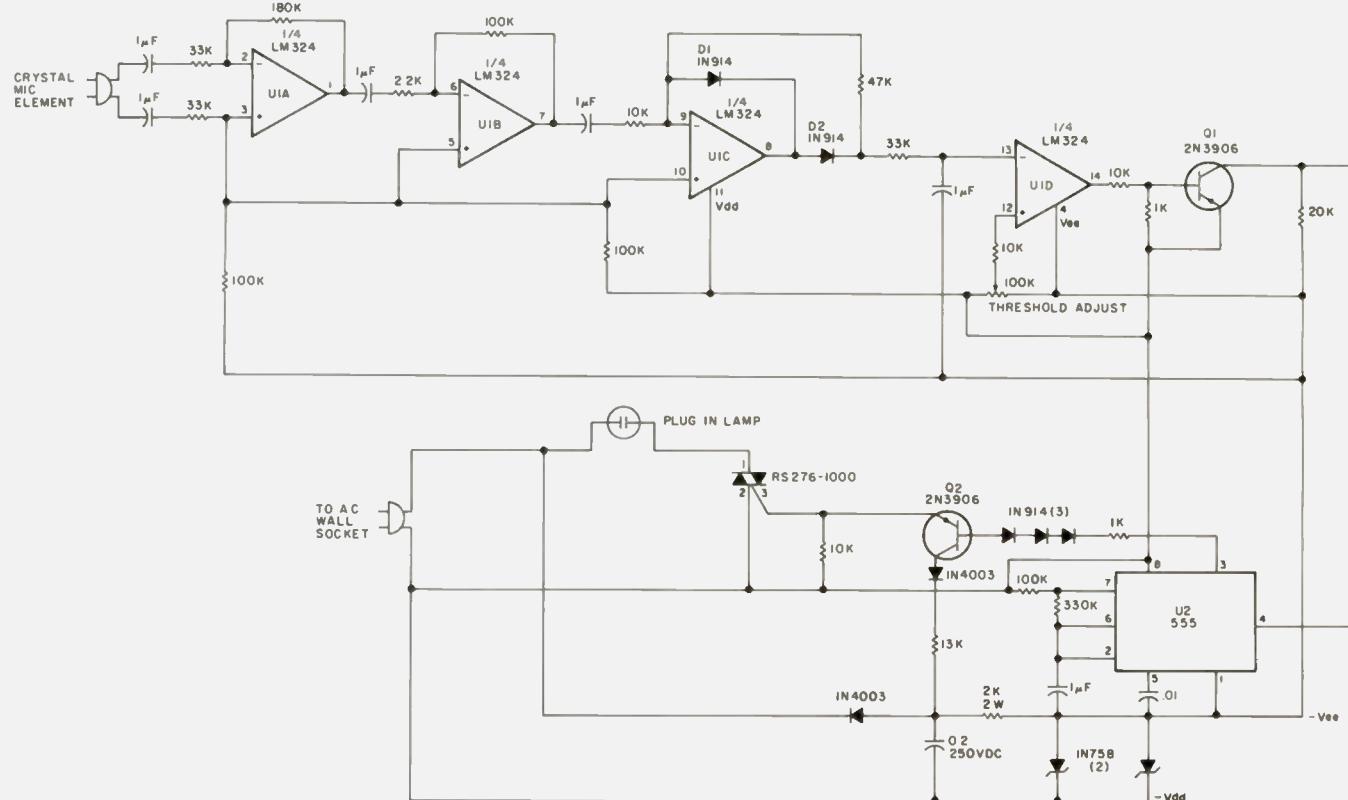


Fig. 1.

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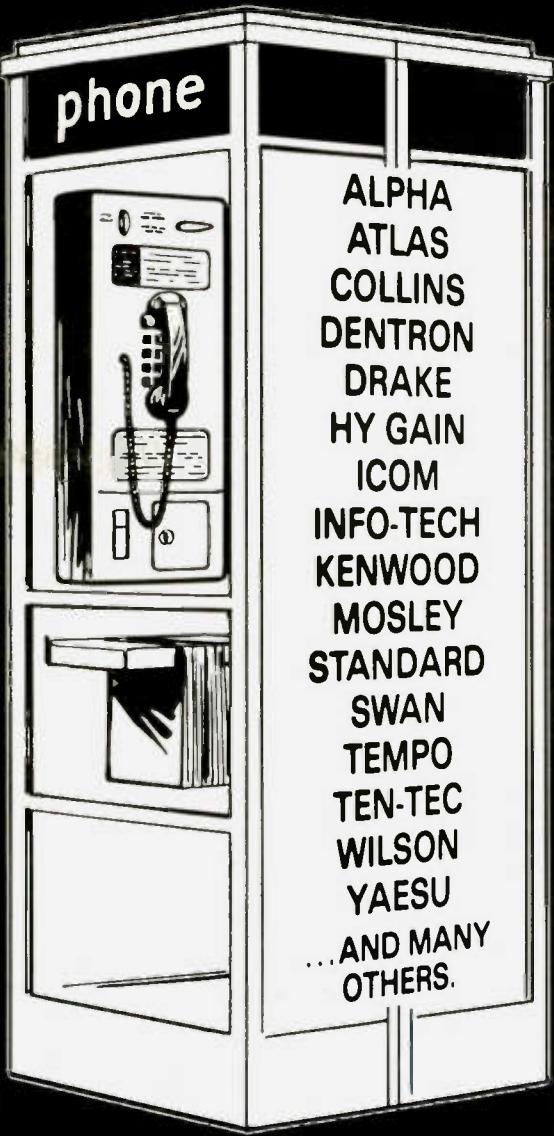
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Morse Converter for DMMs

— super gadget for blind hams

A digital converter which will take the binary coded decimal (BCD) output of a digital multimeter (DMM) and convert it to Morse code would surely be an asset to any blind or partially blind person's workshop.

After graduating from a CREI home study course in electronics engineering technology (minicomputers and microprocessors as the major elective), I had a strong desire to make some practical use of the knowledge.

Shortly after finishing the home study course, I was talking with my dear friend of nine years, W6LZV. He was mention-

ing that a Morse clock could be helpful to him in his daily on-the-air operations. After thinking about a Morse clock versus another type of device which could be converted to a Morse code output, I asked Jean if he could use a digital multimeter with Morse code output. Though Jean is completely blind, he is very active in the building and repairing of electronic equipment. Therefore, it was no surprise to me when he said that the DMM-to-Morse code converter would be more functional.

This article is for the person who wants the challenge of using an instrument which was originally

made for the sighted person. This article can open doors that have inhibited the handicapped person in delving into a world of technical discovery—a world which can be just as satisfying for him as for the sighted person.

While researching what chips might be commercially available to handle the BCD-to-Morse code conversion in a very simplified manner, I ran across the July, 1977, issue of 73 Magazine, which has an article entitled "The Morse Clock." After reading and rereading the article, trying to fully understand the logic, I decided that I would use the article as a starting point for my

design of the DMM-to-Morse code converter.

The digital multimeter used for this prototype project was the Sabtronics 2000 DMM. At first, I had only a schematic of the DMM to work from, but it was apparent that converting the output of the MC14433 A/D converter chip into Morse code would not be too difficult. After building the DMM and seeing how the MC14433 chip worked, I decided that since the position of the decimal point depends only upon the function button and range button selected and upon whether the X10 range switch is used, I would disregard the decimal point. Only the minus sign and the 3½ digits were dealt with.

Features of the DMM-to-Morse Code Converter

The converter was designed to provide the following:

1. Morse code output of the 3½ digits.
 2. Morse code character "m", which signifies that a minus sign is present.
 3. An audible over-range indicator.
 4. A method by which the user can perform "continuity tests."
 5. Three selectable code speeds.
 6. A Morse code zero (0) for a "blanked" first digit.



The converter at work. The DMM indicates -108.5 Ohms. (The minus sign is peculiar to the Ohms reading on this DMM.) When the INI button is pushed, the converter will respond with ---(minus) ----- .

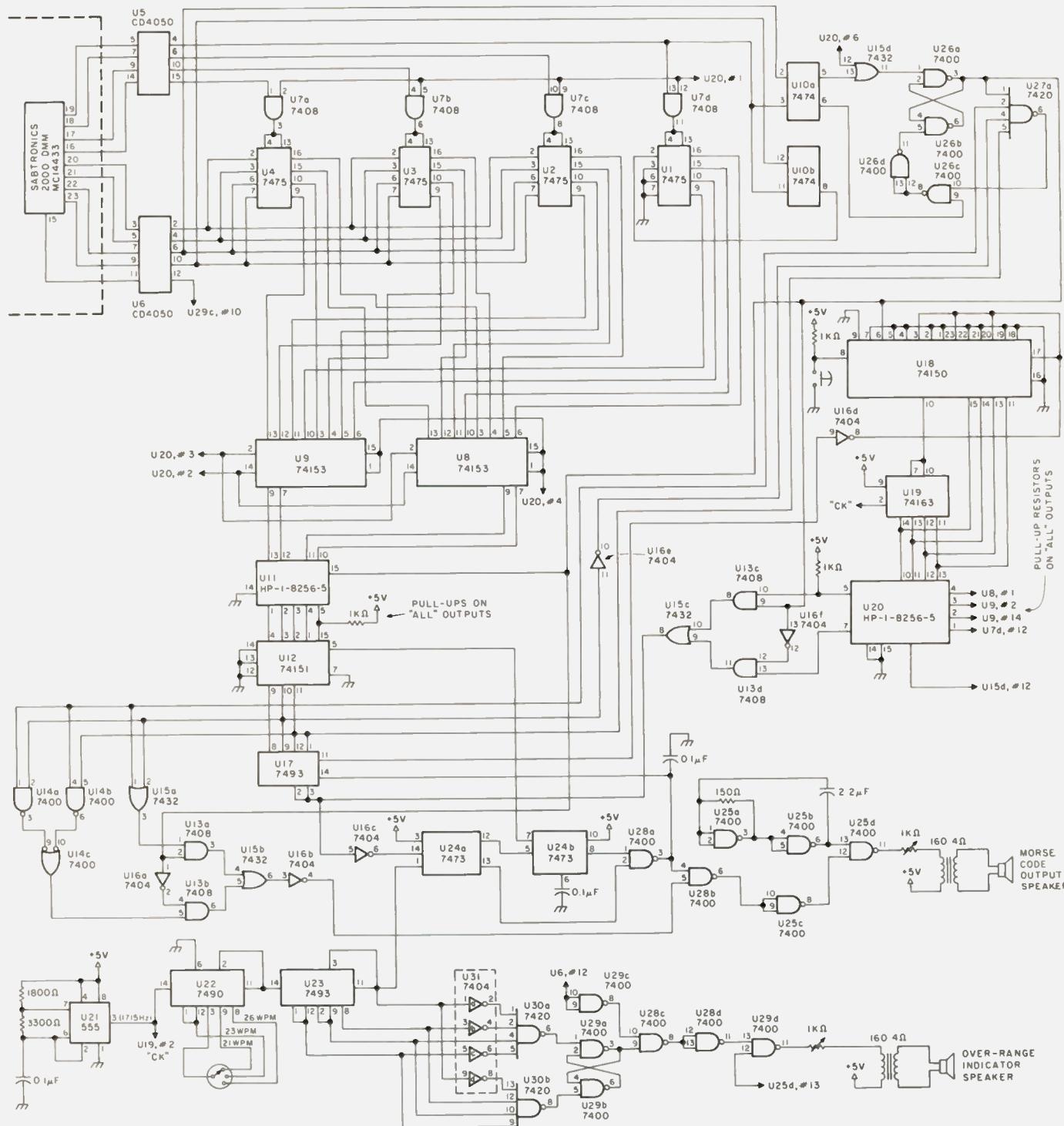


Fig. 1.

This was done because I am not audibly providing the position of the decimal point. By giving the user the contents of all 3½ digits, the mental determination of the decimal point's placement is made easier.

Circuit Description

The MC14433 A/D converter chip (enclosed in the DMM) is the heart of the

converter. The chip has BCD outputs available for the 3½ digits, but they must be demultiplexed "off-chip" by using the display triggering lines. Since the demultiplexing and selection of each digit is not available "on-chip," the chip count of the DMM-to-Morse code converter is higher than for the Morse clock.

Since I have never tack-

led a complete project from the design stage to the building stage, various aspects of the MC14433 chip posed some minor stumbling blocks. After reading the article entitled "Super DVM" (August, 1977, issue of *73 Magazine*), I selected a single 7474 (dual D flip-flop) to demultiplex the $\frac{1}{2}$ (first) digit and the minus sign. The 7474 chip solved the minor

stumbling blocks.

The Minus Sign

The minus sign demultiplexing, sampling, storage, and recovery circuits are made up of U10A, U15D, U26A and B, U26C and D, and U27A (respectively).

When a minus sign is present, the 7400 SR flip-flop (U26A & B) provides the following actions: 1. Selects the appropriate

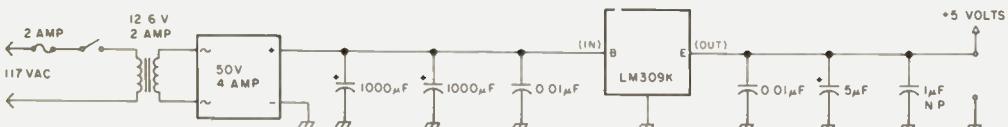


Fig. 2.

State	IRENABLE (+), U13D/pin 13			MINUS SAMPLING, U15D/pin 12			IRENABLE (-), U13C/pin 10			DIGIT LOCK-UP, U7D/pin 12
	I	R	E	N	S	A	M	P	L	
0000	X	X	X	X	X	X	X	X	X	
0001	X	X	O	X	X	X	O	O	O	
0010	X	X	X	O	X	X	O	O	O	
0011	X	X	X	X	O	O	O	O	O	
0100	X	X	X	X	O	O	O	O	O	
0101	O	X	O	O	O	O	O	O	O	
0110	X	X	X	X	O	O	X	O	O	
0111	X	X	X	O	O	O	X	O	O	
1000	O	X	O	O	O	O	X	O	O	
1001	X	X	X	X	X	X	O	O	O	
1010	X	X	X	X	X	X	O	O	O	
1011	O	X	O	O	X	O	O	O	O	
1100	X	X	X	X	X	X	X	O	O	
1101	X	X	X	O	X	X	O	O	O	
1110	O	X	O	O	X	X	O	O	O	
1111	X	X	X	X	O	O	O	O	O	

Table 1. Contents of the "control" ROM (U20).

Remarks

Wait for "next instruction"
Address (-) Minus sign
Send (-) Minus sign
Clear BCD-Morse code converter
Address 1st digit
Send 1st digit
Clear BCD-Morse code converter
Address 2nd digit
Send 2nd digit
Clear BCD-Morse code converter
Address 3rd digit
Send 3rd digit
Clear BCD-Morse code converter
Address 4th digit
Send 4th digit
Clear BCD-Morse code converter

output line (IRENABLE -) from the state-machine ROM; 2. Selects the appropriate logic used for giving the "space" between the Morse code characters; 3. Disables the BCD-to-Morse code ROM. When this ROM is disabled, all of its outputs are driven HIGH, thus providing the needed output for making the Morse code character "m"; 4. Signals the input multiplexer (U18, 74150) that the "m" and the "space" between characters has been sent (no-code condition). Spacing between the Morse character "m" and the first digit is longer than between "each" digit. This characteristic provides a distinguishable pause between the "m" and the start of the first digit.

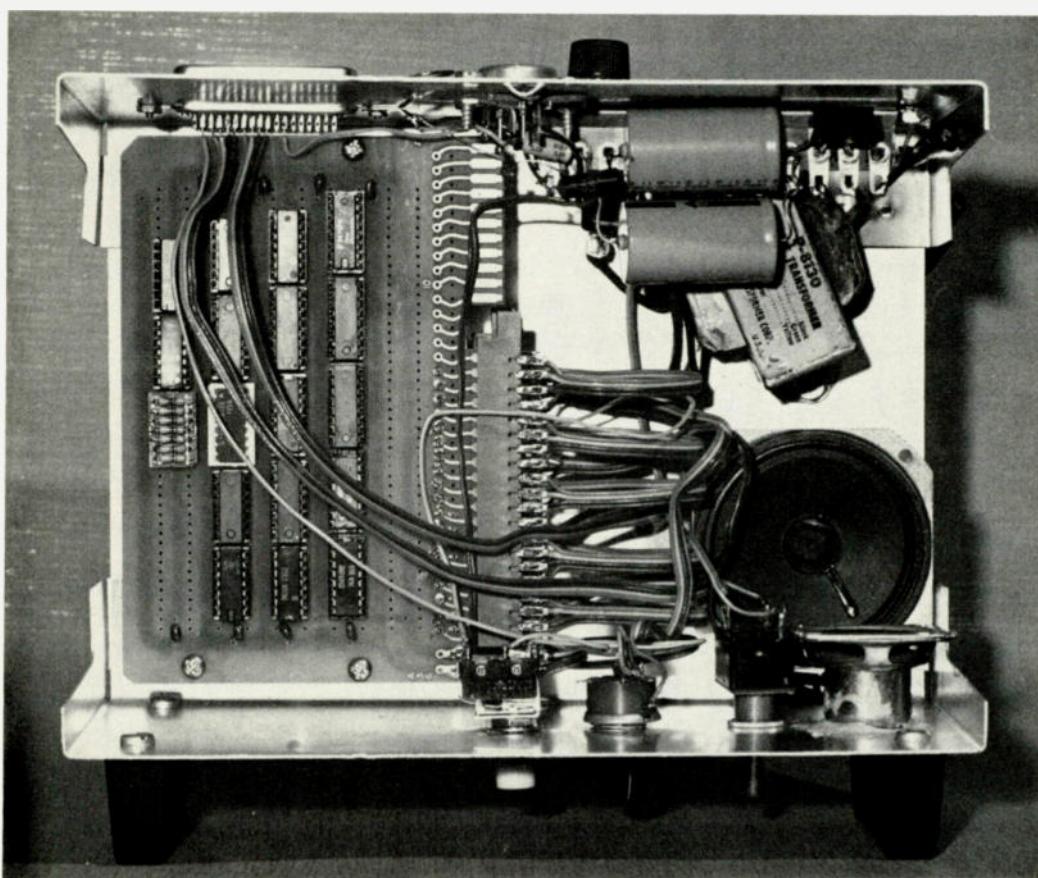
The First Digit

The $\frac{1}{2}$ (first) digit demultiplex and storage circuitry is made up of U10B and U1 (respectively). Since the $\frac{1}{2}$ digit is only a zero (0) or a one (1), I tied the binary 2, 4, and 8 inputs of U1 to ground (binary zero) and let the output of the 7474 dual D flip-flop (U10B) determine whether a zero or a one is showing in the $\frac{1}{2}$ digit's place.

The Over-range Indicator

A 7493 counter (U23), wired as a divide-by-10 counter, is used to provide the counter-states which make up the "ditters" when an over-range condition exists. The 7493 counter-states of 2 and 7 are decoded and driven to a 7400 SR flip-flop, which provides the ditter sensation.

Besides providing the user with an indication of an over-range condition, the over-range indicator circuitry can also be used to provide a continuity test. When the DMM is placed in the OHM function and an OPEN condition exists between the input terminals, all digits are



Top view of completed unit.

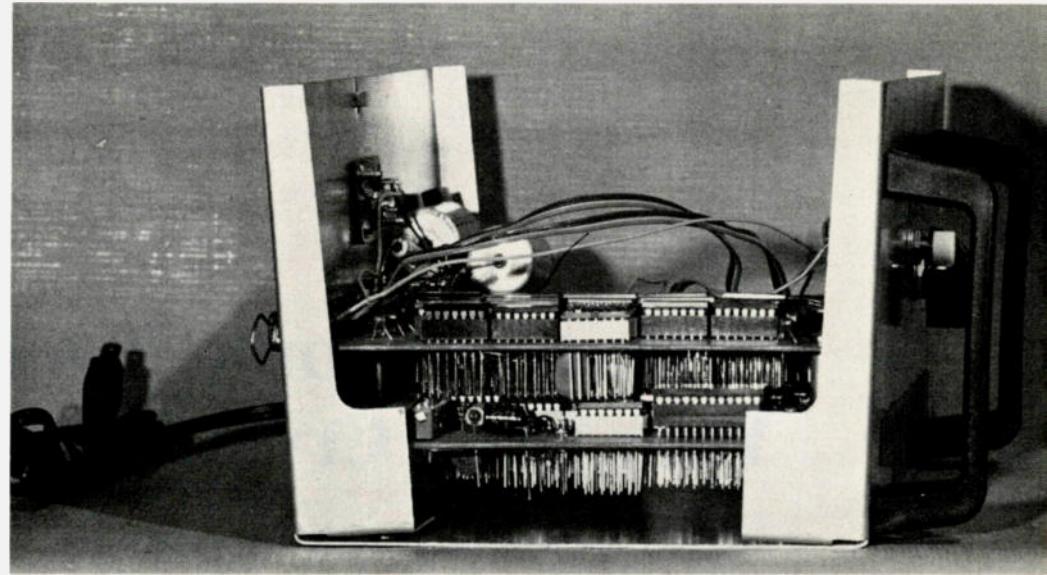
blanked (to the sighted user) and the over-range ditters are HEARD. With this condition in mind, the user takes the test leads and makes the necessary connections for determining if there is an open or a shorted condition. After making the connections, if the ditters persist, an open condition exists (verified by pressing the "initiate" button and hearing "m" 19 99). If the ditters cease, the user either has a shorted condition or some actual value of resistance. A shorted condition can be verified by pressing the "initiate" button and hearing either "m" 0000 or just 0000.

When the function switch is in the OHM position, the user must ignore the "m" (minus sign) which is active due to the design of the DMM. During the OHM function, the minus sign shows up (for the sighted person) to tell the user that the input jack (positive jack) is in fact the negative probe, and that the DMM is in fact turned ON.

Construction/Miscellaneous Notes

Excluding the MC14433 in the DMM, the prototype project has a chip count of 31. All ICs used in the project were of the commonly-available type. All parts were either obtained locally or through various electronic parts catalogues.

I used the wire-wrap technique because it gave the amount of versatility in construction that I wanted. I was able to build each



Side view of completed unit.

functional section bit-by-bit.

Interfacing of the CMOS 14433 with the TTL 7475 and 7408 was achieved by using two CD4050s (hex buffers).

I am sure that there are many alterations which can be made to my present design. These could include feeding the Morse code output and the over-range output through a "summing" op amp amplifier, thereby needing only one speaker. Also, U5 and U6 could be deleted if we dealt with CMOS technology throughout the converter. ■

The cost of the converter parts (including the cabinet and wire-wrap PC boards) was approximately \$120.00, but the cost could be brought down easily to approximately \$60.00 if the builder has a well-stocked junk/spare-parts box and if

he shops around.

My sincere thanks to WB3EVS for his technical assistance, to WA6DLI for his superb photographic shots of the converter, and to WA1MXV and WA5VQK for their very enlightening articles. ■

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2. Tim Ahrens, "Super DVM... Uses the MC14433 and LCD," *73 Magazine*, August, 1977, p. 108.
3. Don Lancaster, *TTL Cookbook*, Sams Publication.

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Table 2. Contents of the "code conversion" ROM (U11).

How to Home-Brew Your Own Crystal Filters

— the series-string method

73 Magazine Staff

One of the prime limitations that discourages many amateurs from pursuing various interesting projects is the cost of good crystal filters. One example is with the construction of rf speech processors or clippers. Clippers are relatively easy to build and very effective, but they require the use of a good SSB filter.

One of the simplest forms yet of a good home brew crystal filter was recently discovered. Hopefully, it will allow many amateurs to pursue some of those projects they have been putting aside because of the expense of commercial filters.

The crystal filter to be described can be built around a group of almost any commonly available MHz-range crystals of the

same frequency. The crystals which are manufactured for TV games or color TV sets are particularly suitable since they are manufactured to close tolerances and are inexpensive. For instance, the 3.579545 MHz crystals for color TV set circuits are readily available for only around \$1.00 because they are mass produced in such quantities. In fact, with a bit of hunting through sales flyers, one often can find them on sale at 6 for \$5.00.

As the title of this article implies, the circuit for the crystal filter is nothing more than a series-connection of the crystals, as shown in Fig. 1. The input and output of the filter are terminated by 1k-Ohm resistors. The trimmer capacitors are used to obtain the overall desired response. The type of response that can be obtained is shown in Fig. 2.

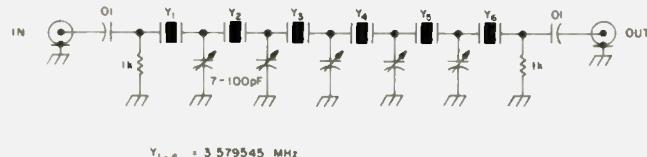


Fig. 1. The series-string crystal filter.

The frequency scale has been expanded somewhat on the graph, so at first glance the filter response may seem rather broad. Actually, it is quite good if one studies the response in a bit more detail. The -60-dB bandwidth is about 8 kHz. The ultimate out-of-passband rejection

could not be measured much beyond -60 dB with the equipment available, but the ultimate rejection is probably -80 dB or better. Just for comparison, the response of a medium-priced (\$40 range) imported filter is shown by the dot-dash lines inside the series-string filter

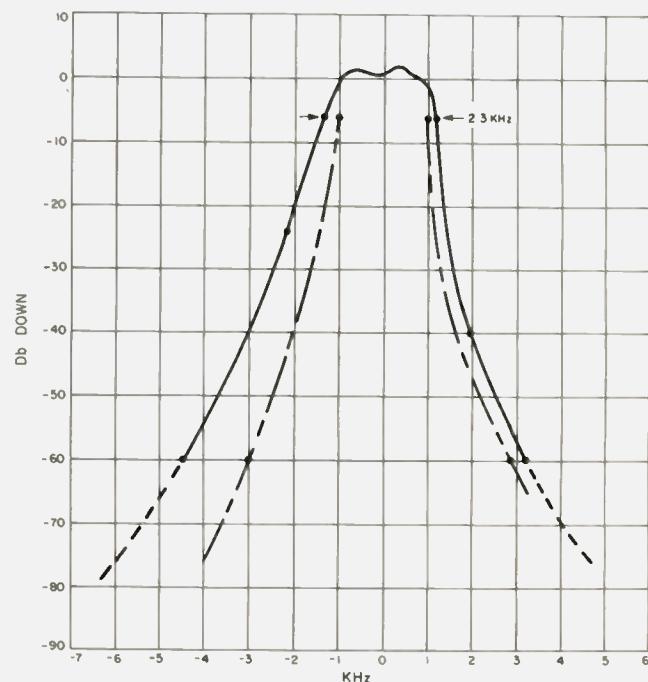


Fig. 2. Response of the series-string crystal filter (outer curve) compared to that of a moderately-priced commercial filter (inner curve).

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response. The commercial filter is better, of course, and one will notice some difference on receiving, but it is a moot point as to whether any difference would be noticed when SSB filtering is considered for transmitting purposes or when rf speech processing equipment is being considered.

In order to achieve the response described, the filter must be constructed carefully, although this is not a tedious task. The filter is constructed basically as shown in Fig. 1, so the input-output terminals are as far apart as possible. Fig. 3 shows the layout used on an approximately $1\frac{1}{2}'' \times 5\frac{1}{4}''$ piece of copperclad (single-side) PC board. There is no need to do any etching unless one just prefers to do this. The isolated-pad type of construction works very well, where a small hole is drilled for each crystal lead or pin (depending on the type of crystal used). The copper around the hole is removed for a small radius with an isolated-pad tool or with a sharp knife. The crystals are simply interconnected, and the few components required are soldered directly to the crystal leads or pins. The input-output cables are connected to any form of isolated terminal post. The trimmer capacitors used are Arco 423 units which have a range of 7-100 pF, but any similar units can be used. In fact, if one doesn't devote a little effort to finding reasonably-priced trimmer capacitors, they can cost more than the crystals!

The adjustment of the five trimmer capacitors will have a great deal of influence on the bandpass shape of the crystal filter. If one has fancy test gear available, such as spectrum display units, the work is, of course, relatively simple. But, even if one

has only relatively simple test equipment—rf generator and a suitable oscilloscope or a receiver—the filter can be adjusted easily enough.

First, set up the filter so the middle trimmer-capacitor is at minimum value, and the other capacitors are at maximum value. Then, as one goes through the passband of the filter with the rf generator, note how the response changes as the center trimmer is increased in value and the other trimmers are decreased. If one keeps a record of the responses for each adjustment of the trimmers, it will not be difficult to discern the pattern that develops. No attempt was made to adjust the trimmers by ear when the filter was used in a receiver, but possibly, with enough patience, it could be done.

The application of the filter to a piece of equipment, or the use of the filter in an accessory item, can follow the scheme shown in Fig. 4. This little setup made use of various International Crystal Company modular mixer and oscillator units to translate the crystal filter response to another i-f frequency that was of interest. However, this setup, or one similar to it that is suitable for the frequencies involved, can be used for many applications.

For instance, one can use it just to improve the i-f selectivity in an older piece of equipment, or those who are a bit more advanced in circuit construction can build a complete outboard rf speech processor for an SSB transmitter, using two crystal filters. (See Fig. 5.) A DSB signal is first generated, and one filter used to produce an SSB signal. The signal is then clipped and passed through the second filter to clean up the distortion products. The product

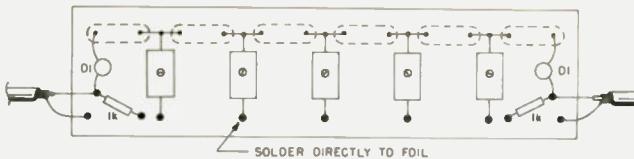


Fig. 3. Layout of a crystal filter board, foil side up. (Crystals are on the non-foil side.)

detector is used to obtain the final af signal. In this application, the two crystal filters should have identical passband responses—but this should not be difficult to achieve if they are constructed and adjusted together. During the adjustment process, one should note where the 20- to 30-dB down points are, on either the upper or lower side of the crystal filters, since either one of the two frequencies can be used as the carrier oscillator frequency.

Construction has been started on such a project, but it is not yet completed. Even using a modular approach as much as possible, by using kit-type modules of the kind shown in Fig. 4, the total cost for the project seems to be coming out in the \$25-\$30 range. That is not bad at all when one considers that it

includes the cost of two crystal filters, and one can pick up several dB of extra intelligibility under QRM conditions. Even if one does not achieve the full theoretical improvement possible with rf clipping, one can easily do better than by using an outboard audio compression or clipping accessory, and at nearly the same cost.

It is possible to use the crystal filter also as part of an adapter to add rf clipping to an SSB transmitter, by connecting it into the SSB generation chain. Unfortunately, one cannot generalize on how this can be done, since the response of the filter in the outboard adapter as compared to the response of the filter in the transmitter, along with the carrier oscillator frequencies, all have to be taken into account for a specific transmitter. ■

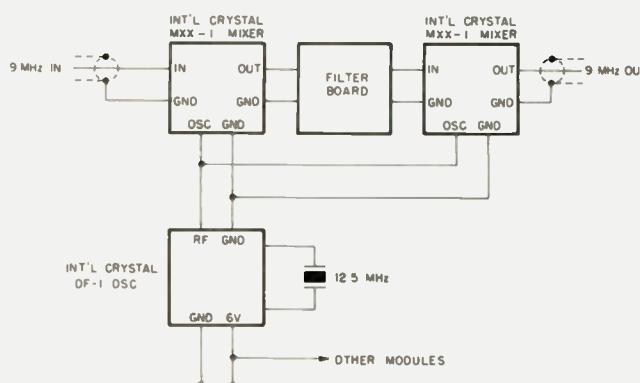


Fig. 4. One example of how the filter can be used in an i-f chain by frequency translation, using readily-available mixer/oscillator modules.

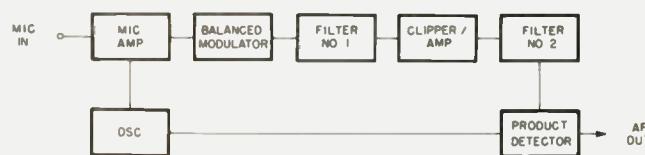


Fig. 5. Block diagram of a separate rf speech processor which now can be built at low cost using two series-string filters.

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Four Bands on a Bamboo Pole

— try a Chinese vertical slanter on 10 through 40

William R. Stocking WØVM
1030 Weidman Road
Manchester MO 63011

Many of us hams like to work DX. My friend, Harry, who lives down the street, had worked many DX stations. I had worked no DX at all! I asked Harry what I could do to work DX.

"Get a long, bamboo fishpole and build a vertical J like mine," he replied. "Get the bottom of the antenna higher than the electric light wires. Get the thing resonant, and you'll work across the big pond."

Harry was my "consulting engineer," and taking his advice usually brought good results. I purchased a 23-foot bamboo fishpole and mounted it high on the house using some lightweight 1 by 2 lumber. A vertical J

antenna was suspended from the top of the fishpole. The matching stub was tuned to frequency, and the 600-Ohm open-wire feedline was attached and adjusted for the best transfer of rf energy into the antenna system. As usual, Harry was right. I got a big thrill the next day when I worked my first European station!

Seeing the big bamboo fishpole, one of my friends called it a "Chinese antenna." When I mentioned this to Harry, he laughed and remarked, "Maybe we can shanghai some DX." Since that time, bamboo fishpoles have been used in most of my vertical antennas, and the "shanghaiing" has been very good. (Fiberglass fishpoles 20 feet long also work well in antenna construction, but they are more expensive and are hard to find.)

This article will present information from some books and a magazine arti-

cle and explain how this information was applied in making a four-band "Chinese vertical slanter" antenna system that is the best non-beam DX antenna that I have ever used.

In the December, 1978, *73 Magazine*, antenna axioms were presented.¹ Among these were: (1) The antenna system should be resonant at the frequency being used. (2) The rf from the antenna must be effectively put into the antenna. (3) There is no substitute for height; the higher the antenna, the better. (4) At easy-to-attain heights, vertical antennas have lower angles of radiation than do horizontal antennas. This makes them good for DX.

Because of their lower angles of radiation, vertical antennas have been known to outperform horizontal beams in making DX contacts, especially in cases where the horizontal beam was not any higher off the ground than the

center of the vertical antenna. The value of low-angle radiation in working DX is beautifully explained in Capt. Paul Lee's book in Chapter 1, "Optimum Design For DX."² This chapter also contains other valuable information useful to antenna designers.

Vertical antennas longer than a quarter wavelength radiate rf at lower angles to the horizon than do quarter-wavelength ground-plane antennas. The angles of a quarter-wavelength ground-plane antenna are from 10 to 55 degrees. For a 3/8-wavelength vertical antenna, the angles are from 8 to 40 degrees. A half-wavelength vertical's angles are from 5 to 35 degrees. The angles of a 5/8-wavelength vertical are from 3 to 27 degrees.³ For 3/4-wavelength verticals, "there is some slight deterioration of the low angle vertical pattern, but this is not serious."³

Vertical antennas longer

than a quarter wavelength also have power gain as compared with a quarter-wavelength ground-plane antenna.⁴ A half-wavelength vertical antenna has a power gain of 1.8 dB as compared with a quarter-wavelength ground-plane antenna. A 5/8-wavelength vertical antenna has a power gain of 3 dB as compared with a quarter-wavelength ground-plane antenna. One can surmise also that a 3/4-wavelength antenna has some power gain as compared with the ground-plane antenna.

From this information, it is clear that vertical antenna lengths between 1/4 and 3/4 wavelengths will be better for making DX contacts than the popular coax-fed ground-plane antenna. To use these ideas in a multiband vertical antenna, tuned feeders must be used to take the rf from the transmitter to the radiating wires. When tuned feeders are used, balanced antennas can be built with wires almost any length, provided that the length each side of the tuned feeders is the same. Furthermore, such an antenna can be "loaded up" and used on several bands.

My friend Harry was somewhat impressed with this information. He suggested, "Why don't you make a chart comparing a tuned-feeder vertical with the popular trap vertical that is about 22 feet tall? After you get that one figured out, figure out what length of wires would make the best all-around DX antenna for 40, 20, 15, and 10 meters."

My "figuring" was as follows. For all practical purposes, the trap vertical is a quarter-wavelength ground-plane antenna on all bands, with the highest frequency radiators closest to the ground. As such, the angles of radiation with respect to the horizon are from 10 to 55 degrees on

all bands. Except for the 40-meter band, on which all 21½ feet radiate, the trap vertical does not have its radiating antennas as high in the air as does a 22-foot "vertical slanter." (Axiom No. 3: "There is no substitute for height.")

With the vertical slanter, all 22 feet radiate on all bands. With respect to power gain and angles of radiation, the vertical slanter has the following as compared with the ¼-wavelength ground-plane antenna with its 10 to 55 degrees angles of radiation on all bands: On 10 meters, the vertical element is 5/8 wavelength with a gain of 3 dB and angles of 3 to 27 degrees. On 15 meters, the vertical element is 1/2 wavelength with a gain of 1.8 dB and angles of 5 to 35 degrees. On 20 meters, the vertical element is a bit shorter than 3/8 wavelength, would have some gain as compared with the trap vertical, and the angles of radiation would be somewhat lower than those of the trap vertical.

With a ground-plane antenna using 4 equally-spaced radials, the radials do not radiate because the fields of the radials cancel out each other. Drooping the radials increases the gain of the ground-plane antenna.⁵ Therefore, it is logical that a vertical-slanted tuned doublet, or a one-radial tuned ground-plane antenna might have some gain compared with a quarter-wavelength ground-plane antenna because there are no other radials to cancel radiation from the lower half of the antenna.⁶

Another great advantage of a vertical-slanted tuned doublet is that it can be tuned to exact resonance at any frequency, phone or CW, on any band, 10, 15, 20, or 40 meters. This means that the antenna will load up equally well on phone and CW in all of

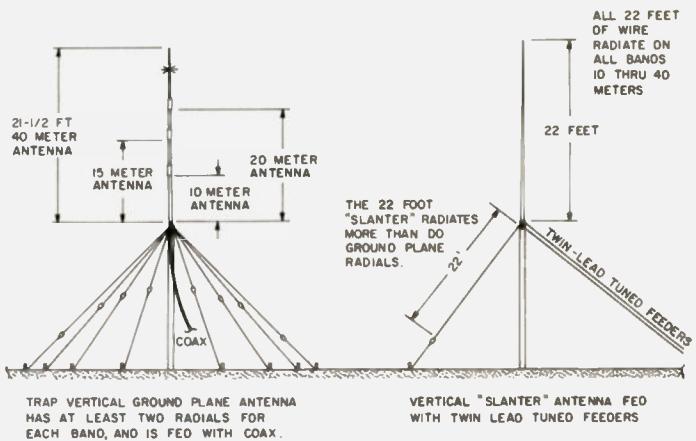


Fig. 1. Typical trap vertical and "vertical-slanted" four-band antenna fed with tuned feeders. The bottom of each vertical element is 20 feet above the ground.

these bands.

Harry made this suggestion: "When you figure the best wire lengths for your four-band DX vertical slanted, make the wires as long as you can without degrading the performance on 10 meters. Of course, if you want better performance on 40 and 20 meters and don't care about 10 meters, you could make your wires as long as possible without degrading the 15-meter performance."

With this in mind, I chose 26 feet as the wire-length for the best performance on 10, 15, 20, and 40 meters. Twenty-six feet is 3/4 wavelength for 10 meters. If I did not care to work 10 meters, I would have chosen 33 feet, 3/4 wavelength for 15 meters.

The gain and angles of radiation for the 26-foot wire vertical-slanted antenna are conservatively stated as follows: 10 meters (3/4 wavelength), "guesstimated" 2.5 dB, with angles 5 to 35 degrees; 15 meters (5/8 wavelength), 3 dB, with angles 3 to 27 degrees; 20 meters (.39 wavelength), "guesstimated" 1.4 dB, with angles somewhat less than the 10-50 degrees of a quarter-wavelength ground-plane antenna. On 40 meters, 26-foot-long wires would work better than 22-foot-long wires.

Since in theory a vertical-slanted antenna fed with tuned feeders seemed to be much better than a four-band trap vertical, I decided to build one and try it out. I had an old, used piece of no. 12 Romex, 25' 8" long. This was very close to 26 feet, so I took out the two wires and used them in the antenna. A bicycle whip taped to a bamboo fishpole clamped to a small sailboat mast supported by two 8' two-by-fours clamped to a fence post was used to support the vertical part of the antenna. The black wire was taped to the supporting structure as the vertical element. (The black color helped to absorb heat and melt the ice after an ice storm.) The white wire was used as the slanted which was drooped down and out towards the south. 300-Ohm twinlead was used for the tuned feedline. The "Chinese vertical slanted" antenna looked interesting, but would it really work?

I thought that 40 meters would be the band on which the antenna would be least effective. (If thought of as a one-radial tuned ground-plane antenna, the vertical element is less than a quarter wavelength long. If thought of as a center-fed tuned

doublet, the total antenna length is a shortened half wavelength.)

On Friday, December 8, 1978, I decided to try out the antenna on 40 meters (worst band first!). The first station called, DJ4IT (West Germany), came back with a 579 report. At the end of this QSO, SP7EJS (Poland) was calling and reported a 559 signal when contact was made. After supper, LZ1KUF (Bulgaria) was called, but he did not come back. (I was not sure whether or not he had called CQ.) The next two calls resulted in QSOs. DL7PR (West Germany) reported 569, and YU1QFX (Yugoslavia) reported 599. (I suspect that this report was exaggerated for the sake of international goodwill. Hi!) These QSOs convinced me that the antenna worked well on 40 meters. Since that time, with very little time on the air, many

DX QSOs have been made on 20, 15, and 10 meters. On several occasions, the DX station chose to answer me instead of one of the other stations calling him. This "Chinese vertical slanter" antenna is the best non-beam antenna that I have ever had. After a storm took down my quad, I could continue to have fun working DX.

Remember that a vertical slanter fed with tuned feeders will load up equally well on both CW and phone frequencies. The above-described antenna even loaded up and made a few contacts on 80 meters. However, the wires are much too short for good performance on the 80-meter band.

For even better performance, a system of radials each about 35 feet long, buried in the ground, could be added to the antenna system. The center of the

radials should be under the vertical element. The more radials, the better. However, the antenna works well without radials as the above information indicates.

I have no trap vertical with which to compare the "Chinese vertical slanter" using on-the-air contacts. Certainly DX can be worked using trap verticals and other coax-fed ground-plane antennas. However, antennas with more gain, lower angles of radiation, and more effective feed systems will certainly outperform any comparable coax-fed ground-plane antenna when it comes to making DX contacts.

I think you'll enjoy the "Chinese vertical slanter antenna." Build one, use it, and find out for yourself how good a non-beam antenna can be for working DX. ■

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- "Tuned Feeders and Other Good Stuff," William R. Stocking W0VM, *73 Magazine*, December, 1978, p. 118.
- Capt. Paul H. Lee (USNR) K6TS, *The Amateur Radio Vertical Antenna Handbook*, p. 13.
- Edward M. Noll, *73 Vertical, Beam, and Triangle Antennas*, p. 35.
- William I. Orr W6SAI and Stuart D. Cowan W2LX, *Simple, Low-Cost Wire Antennas for Radio Amateurs*, p. 43.
- William I. Orr W6SAI and Stuart D. Cowan W2LX, *The Radio Amateur Antenna Handbook*, pp. 88 and 92.
- This idea does not take into consideration ground losses. It might be that four drooping radials, each the same length as the vertical element, would enhance the performance of the antenna because there would be less ground loss. Radials buried in the ground under the vertical-slanted antenna might improve its performance, and they are suggested near the end of the article.



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The Triton IV Goes QRP

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C. Stewart Gillmor W1FK
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My last tube rig (I sold it 20 years ago) had something like 1400 volts at 300 mA on the plates of the final tubes. It amuses me, while getting good reports on 20 meters, to glance at a meter and see that I am running 3½ volts on the finals of my Triton IV! Basically, this is a good way to run the popular Ten-Tec Triton transceiver efficiently at very low power input levels. In this article, I will show how you can modify the Triton IV for such operation and yet drill no holes and make no changes which could not easily be redone to put the rig back into original specifications. This idea

originated with Mac Harper W1FYM, who made the same modifications to his Triton IV and suggested I modify mine so that our two rigs could form the basis for the portable QRP station at W1FYM/1 for Field Days 1977 and 1978. The Field Day setup was battery-powered, including our computer logging and TV monitor equipment, and it has performed quite well through 2 Field Days. But the computerized QRP Field Day station is another story . . . one that W1FYM must write.

The modifications described here allow the Triton IV not only to operate very efficiently at 10 Watts input for Field Day, but also they allow such operation any time you wish, merely with the changing of the 6-connec-

tor power plug to the transceiver. The design idea could easily be altered to suit other solid-state rigs as well. Now you may say, "But my solid-state rig is easily capable of QRP operation as is. I just back down the drive to the finals until I'm drawing about 0.8 Amps at 12 volts and I'm running about 10 Watts input." "Ah, yes," I reply, "10 Watts input, but very little output!" Instead of 1 or 2 Watts out, your rig could be putting 6 or 7 Watts out for the same 10 Watts input if you ran the finals more efficiently. A good way to do this is to run them at much lower voltage. A very satisfactory choice for the 5-to-10-Watt power class is about 3.5 volts. The necessary modifications for this are described in two parts: modifications to the Triton IV and construction of an integrated circuit variable voltage regulator.

Modifying the Triton IV

Basically, what we want to do here is allow the driver and final amplifier transistors to have their own voltage feed separate from that which feeds the rest of the transceiver and auxiliary equipment. Ten-Tec uses a 6-connector

power plug and socket to run power leads and ac switching from the model 252 and 262 power supplies to the transceiver. Pins 1 and 2 are for ac switching and need not be altered. Pins 3 and 4 are grounded and likewise may be left untouched. Pins 5 and 6 are for +12 V dc. This pair is originally soldered together at the power plug to reduce pin resistance. To modify the Triton IV, proceed as follows (refer also to Fig. 1): Separate pins 5 and 6 at the power plug at the left rear of the Triton IV. Resolder the red #14 stranded wire onto pin 6 alone. Follow this red wire to its other end, through the AGC 20-Amp fuse to the cathode end of the 1N3492R diode (diode D1), at the 3-terminal strip. Disconnect this red #14 wire from diode D1 and run it in a 2-turn loop through the ferrite bead directly to the "+12 Final" terminal on the back of the final amplifier metal box; solder it securely. The "+12 Final" terminal is the large one with the solder lug. What you have now done is make pin 6 on the power plug of the Triton dedicated to providing dc solely to the driver and final amplifier transistors. For

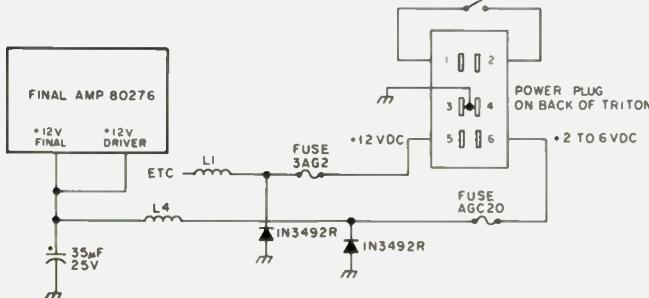


Fig. 1. Schematic of Triton IV as modified for efficient QRP operation.

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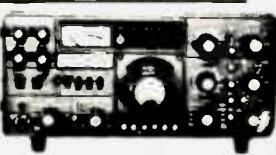
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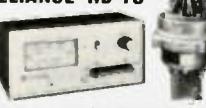
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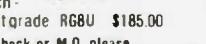
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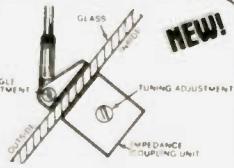
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safety, in case you mistakenly reverse your battery polarity, it might be wise to buy another 1N3492R high-power diode and install it across the line to ground in reverse-bias polarity position. This can be mounted on the opposite end of the 3-terminal strip from the other 1N3492R diode. Solder the cathode end (marked with a bar) to the #14 red wire going to the final amplifier transistors and the anode end to ground. Then, should you goof and reverse the battery leads (such as happened at 3:00 am in the middle of Field Day to yours truly), the diode will conduct, blow the 20-Amp fuse, and save your rig.

Now, take a new (say, orange-colored) piece of #14 stranded insulated wire and solder it to pin 5 of the Triton power plug. Push it through one of the cable holes in the metal shielding around the low-pass filter section and leave a few inches extra. Now take a small screw-mounting 3AG fuse holder and mount this to the outside of the low-pass filter shield. There is a screw very near coil L4 on the low-pass filter board which holds the board to the shield. The fuse holder can be mounted very easily on the back side of this aluminum shield. No holes need be drilled. Note: The toroid-core coil, L4, on the low-pass shield board is not the same coil as "L4" in my Fig. 1. The latter is a ferrite bead through which the final dc supply line is wound for 2 turns of rf shielding.

Now that you have mounted the 3AG fuse holder, connect your orange wire to one end and install a small, 2-Ampere fuse in the holder. From the other end of the fuse holder, run the same size orange wire along parallel

to the back of the final-amplifier metal box until it reaches the 3-terminal strip where diode D1 is mounted. Solder the orange wire to the near end of the terminal strip; this is the same point where the cathode end of diode D1 and also 2 small #24 red wires are soldered. You have now completed the Triton modification. The orange wire from pin 5 of the power plug runs through a 2-Amp fuse and continues on to supply +12 V dc to all of the rig except the driver and final stages. Your regular 6-connector power socket from your model 252 or 262 power supply will work the same as before except that now you have the option to supply pins 5 and 6 on the power plug with different dc voltages if you choose. A way to supply different voltages is described below.

Variable-Voltage Regulation Using an IC Regulator

All you really need to do to supply the Triton final transistors with low voltage for highly efficient QRP operation is provide dc at from 2 to 6 volts at anywhere from 1 to 5 Amps. If you already have a small multi-voltage dc supply, fine. Or you can consult one of the ham handbooks and find a circuit which supplies dc at only a single voltage level. This would work perfectly well, for you can vary power input of the Triton using the drive control. I will describe here a variable-voltage regulated supply which provides 2 to 6 volts out and easily handles 5 Amperes (see Fig. 2).

This was constructed in W1FYM's shack in one evening using some purchased and some scrap parts. The circuit was mounted on PC board. I purchased a heat sink somewhat larger than necessary, and the pass transistor (HEPS7004) is rat-

ed at 15 Amps! Since the case of this large power transistor is not at ground potential but at +12 V dc, we mounted the circuit board on the heat sink, which in turn was mounted on 2 insulating pieces of nylon scrap. The μ A723 IC was mounted in a socket rather than being soldered. I always avoid soldering IC chips, if possible, because everybody ends up frying a chip now and then and it's a lot easier to test or change a chip which is socket-mounted.

Ten-Tec sells a small accessory 625-mV meter (ammeter model 207) which plugs into the various power supplies of their 250 and 260 series, using a phono plug. We wanted to monitor our current input during Field Day and other QRP operation but didn't want to buy or permanently install a meter, so we put a phono jack on the board and wound a wire shunt of about 0.7 Ohms value so that the meter reads 5 Amps full scale. Of course, you could use a different meter movement with a different shunt value. The 1k pot is an Ohmite ANP 102K or similar and its value, as well as that of the 330-Ohm $\frac{1}{2}$ -Watt resistor, is not critical. The IC used by us was a Fairchild μ A723, but equivalents are available from firms including Motorola (MC1723), National (LM723), and Signetics (μ A723). Q1 is a

1-Ampere-rated pass transistor, Texas Instruments TIP29B or equivalent; Q2 is the 15-Ampere-rated HEPS7004. We used #12 stranded copper wire with large battery clips to supply the 12 V dc from the battery to the regulator power plug (pin 5 and pins 3-4). I would recommend using #14 or #16 stranded copper wire from the emitter of Q2 to the shunt and on to pin 6 of the power socket.

The construction and use of the regulator is straightforward. The 1k pot can easily be adjusted for output voltages from about 2 to 6 volts. W1FYM has had only to replace one fried IC chip in a year and a half of occasional use and I have had absolutely no troubles at all. Believe it or not, I've often had 579 and 599 reports on both CW and SSB into Europe and South America on 20 meters. Those reports aren't due only to the Triton IV QRP...I use long vee-beams 6 to 8 wavelengths on a leg and multi-element wire collinear beams at my QTH. Still, I've described a fairly-easy-to-make modification to the Triton IV and the construction of a modest voltage regulator which not only makes for a competitive Field Day rig, but also allows one to use a medium-powered solid-state rig such as the Triton in highly efficient QRP operation. ■

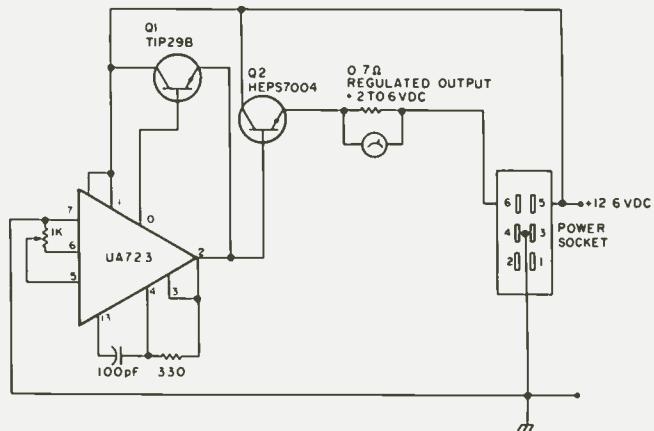


Fig. 2. QRP voltage regulator for the Triton IV.



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Experimenter's Corner: The MM5369N

— how many uses can you find
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Charles B. Patten
4641-4 Chicago Drive
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Would you like a simple, versatile, crystal-controlled oscillator? Would you like it to operate from below 500 kHz to over 16 MHz, with rich harmonic content, simply by changing crystals? Too good to be true? Read on, it gets better!

The MM5369N—see Fig. 1(a) and (b)—is a CMOS integrated circuit that is configured to provide a crystal oscillator (you supply the crystal) with a buffered output (maximum avail-

able load about 500 uA) and an internally-connected divide-by-59659 (or 215.864452175) divider chain which provides, in the manufacturer's intended use, a 60 Hz output for dc-operated digital clocks when used with an inexpensive 3.579545 MHz crystal. This, and numerous other uses, will be mentioned as we proceed.

How It Works

The crystal oscillator is biased by a crystal and RC network (Fig. 2) consisting of R1, C1, and C2. R1 is nominally 20 megohms to bias the oscillator stage for class A operation.¹ Capacitor C1 provides the nec-

essary parallel capacitive load, which, with R1, causes the oscillator to start reliably. C2, in parallel across X1-C1, permits the oscillator output to be trimmed to an accuracy of ± 100 ppm of the desired operating frequency with an accuracy of ± 2 ppm obtainable with careful component selection.

The buffered oscillator output frequency is available at pin 7 of the IC which gives you a convenient point to monitor the oscillator output without loading, while trimming to the desired operating frequency. The buffered output is at CMOS levels, and for that reason must drive only CMOS-compatible circuits. If other than CMOS levels are required, it will

be necessary to provide the needed digital or analog buffers. Several CMOS-to-digital-level buffers are available from many of the suppliers listed in the advertisements in this magazine; additionally, all parts called for in this article are available from Surplus Electronics, 9600 Baltimore Blvd. (rear), College Park MD 20740. The CMOS-to-analog-level buffer is not available as a single chip, but may be constructed in the form of a one- or two-transistor amplifier to allow the oscillator to be used in just about any circuit configuration you can imagine. These can include local oscillators, QRP transmitters, frequency counters, and so on. Note that the output

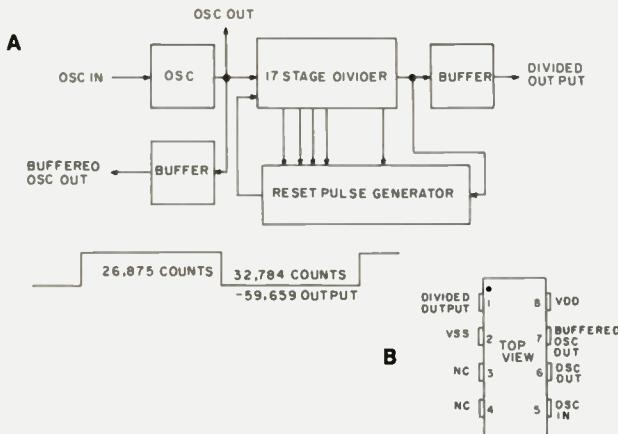


Fig. 1. (a) MM5369N block diagram and timing.¹ (b) MM5369N pinout.¹

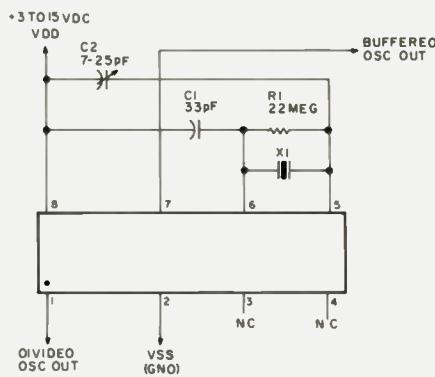


Fig. 2. Basic operating circuit.

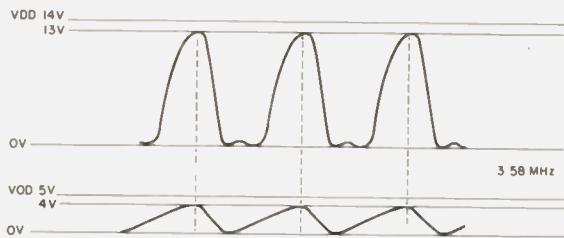


Fig. 3. Output waveforms taken at 3.58 MHz from pin 7. (Note loss of rise and fall times at very reduced Vdd.)

load on the buffered oscillators output must be kept below 500 uA to avoid exceeding the manufacturer's specifications.

The divider section in the standard MM5369N is a mask-programmed, 17-stage divider set to divide the input frequency by 59659. This is accomplished in conjunction with a six-input reset pulse generator which, by virtue of its mask programming, sets the internal divisor length. The MM5369 is also available with custom mask programming. Should you have the need and the bucks (kilo, no doubt), National Semiconductor would, most likely, happily mask-program a few thousand for you with any divisor from 10000 to 98000. The divided output duty cycle stays very close to 50% for a V_{ss} of 8-15 volts and can drop as low as 20% for a V_{ss} of 3 volts at the same frequency. The current drain is nominally 2.4 mA for frequencies of 500 kHz to about 6 MHz, with up to 20 mA drawn for frequencies up to 16 MHz.

Operational Considerations

The MM5369N is a versatile crystal oscillator chip requiring a bare minimum of supporting parts for proper operation. The oscillator is very tolerant of supply voltage variations, and the only noticeable effect of an abnormal voltage change is a possible

Stand-Alone Crystal-Controlled Calibrator

So far you have been presented with a bunch of

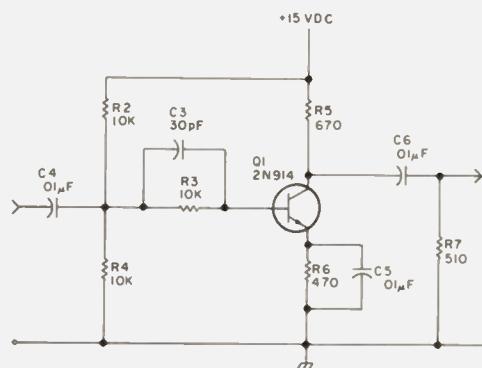


Fig. 4. Modified buffer amplifier for CMOS-to-analog level translation.³

theory and "how to" and little else. Here's your chance to include all of the above plus a little more and have a reliable test instrument standard for your lab. First dig out that old 1- or 10-MHz rock you bought at the last hamfest. We're going to combine it with the MM5369N, a TTL buffer amp, a few ICs, and end up

with a crystal calibrator with multiple divided outputs. It will serve you well when the time comes to calibrate your scope, and, in between, serve as the timebase for a frequency counter, if you wish. The output of pin 7 on the MM5369N (Fig. 5) is fed to the TTL buffer amp and then to the first 7490; each

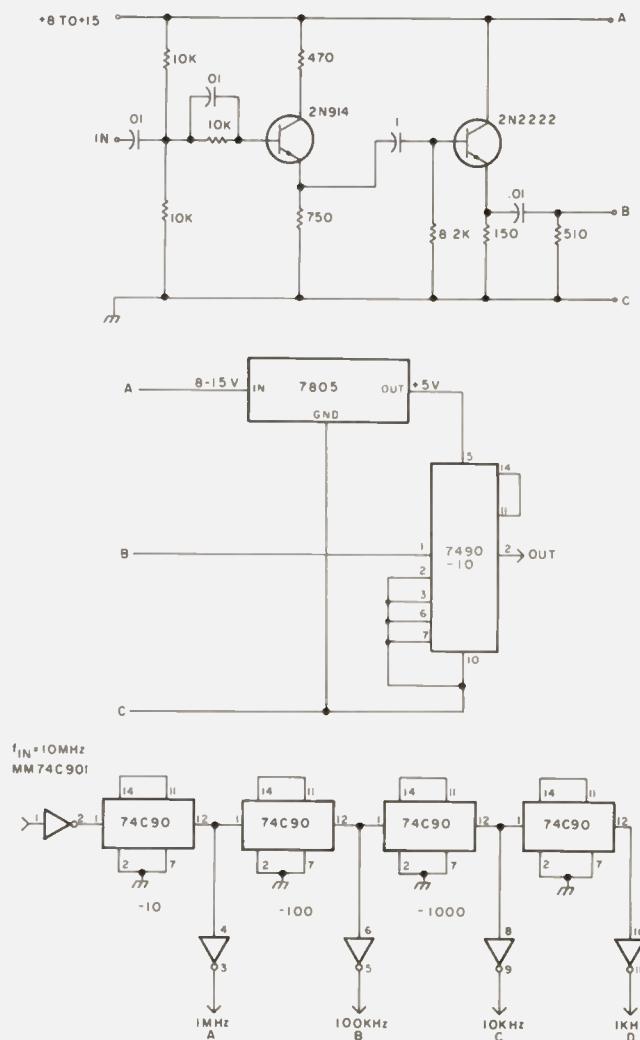


Fig 5. TTL digital interface and divider chain.

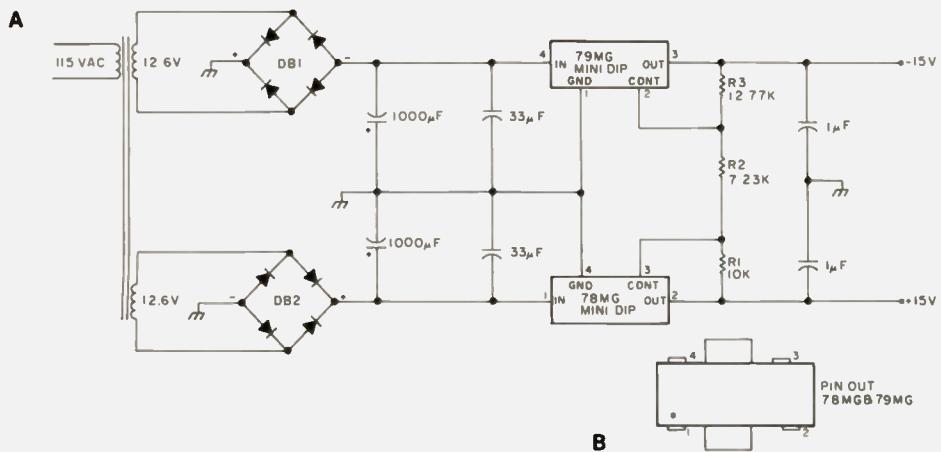


Fig. 6. (a) UA78MG and UA79MG power supply. The negative portion of the supply is not required for this project. It is provided to allow the user greater latitude in using the power supply in other projects as well. (b) UA78MG and UA79MG pinout.

divided signal is then fed to the following IC stage in turn. At each 7490 output, the signal is brought out to a TTL buffer chip where it is available for use as you please. At approximately 100 kHz you may wish to substitute 74C90s for the 7490s to keep the current drain to a minimum.

If you choose this route you should use 74C901 or 74C902 CMOS-to-TTL buffers. These outputs are fed, in turn, to a rotary switch

and from there to the output. There is also a signal provided that has been divided by 59659 that is available at pin 1 of the MM5369N. This also can be buffered and used as a known signal source for a press-to-test function if you use this circuit for a frequency counter time-base. You will note that the TTL buffer is a two-stage amplifier driving a conventional $7490 \div 10$ IC. This is necessary as it still is diffi-

cult to find 74C90s that operate at 16 MHz. So far we have touched on the parts required to have a crystal oscillator but we don't have a source of power to run it—yet.

The Power Source

The power source—see Fig. 6(a) and (b)—is a simple voltage-regulated source that will help eliminate the aforementioned "abnormal voltage drop." It consists of the usual transformer, full-wave diode bridge, and capacitor arrangement followed by a really neat voltage regulator chip, the UA78MG, made by Fairchild Semiconductor. The UA78MG is a 4-terminal positive voltage regulator (UA79MG is the negative version) which is adjustable from 5 to 30 volts at a current of 500 mA. This versatile device can be used with a series pass transistor for higher output currents when needed.² For those who are interested in the gory details of how to design power supplies with this chip, a few of the basic rules I have found to be useful are set forth below:

- 1) Set the control current through R1, R2, and R3 to 1 mA.

- 2) The UA78MG control voltage is 5 V, the UA79MG is 2.23 V. $R_2 = 7.23/1 \text{ mA} = 7.23\text{k Ohms}$. If the power

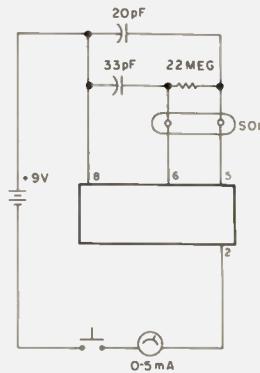


Fig. 7. Crystal checker schematic.

supply is just plus or just minus, then that respective leg is referenced to ground, and R2 is the control voltage divided by the control current.

- 3) The value of R1 and R3 is found by $R_x = V_{out}(\text{desired}) - V_{control}/I_{control}$.

- 4) If you prefer an adjustable supply, R1 and R3 can be replaced by trim pots and their respective voltages set independently.

- 5) Include a .33-μF input and a .1-μF output capacitor for each circuit to improve the circuit's output transient response time.

If you wish, a 25-volt center-tapped transformer may be modified to provide the dual output voltages called for by Fig. 6. Take a sharp knife and carefully follow the c-t lead through the paper until you find the point where the lead is soldered to the output winding. When you find this point, carefully unsolder and split the two winding wires, attach separate output wires, and cover the connections with the proper size heat shrink or tape. Re-wrap the transformer with tape, check for shorts—and you're done.

Other Nonstandard Uses

Other nonconventional uses include a portable crystal checker (Fig. 7). Just add a 0.5-mA meter, and replace the power supply with a 9-volt transistor battery. Note and mark the

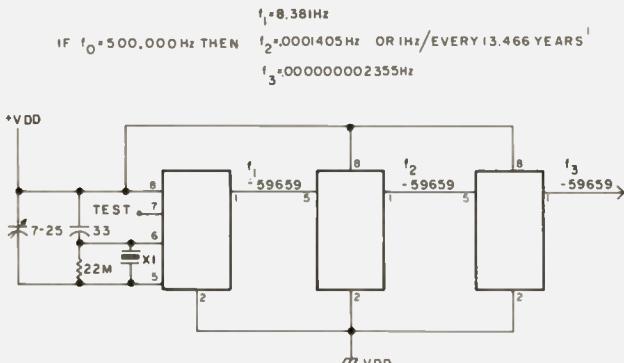


Fig. 8. Long-interval timer-option schematic.

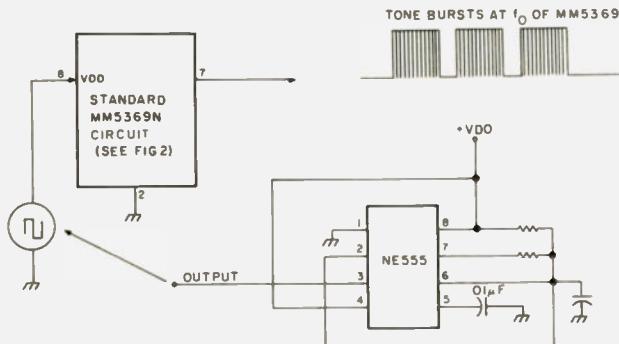


Fig. 9. MM5369N tone or frequency burst schematic and block diagram.

current drawn without a crystal in the circuit and mark this on the face of the meter. Crystals that operate at 5 MHz and below will cause a current drop from the mark, while 7 MHz and above will cause the current to increase. If there is no change in the amount of current drawn, then the crystal is open, or very close to 6 MHz.

The MM5369N also can be used as a long-interval timer. You can cascade the MM5369N without using the active oscillator part of the IC by feeding the output of the first oscillator (pin 1) to the input of the next oscillator (pin 5). (See Fig. 8.) By this scheme, you select the desired output period and multiply by 59659 or multiple thereof, and find the inverse to get the operating frequency needed. If you wish to have 1 Hz every 1.88 years you have to use an operating frequency of 3.579545 MHz and divide it through 3 MM5369N dividers to obtain the desired output.

Another option is to use the MM5369N as part of a tone or frequency burst oscillator (Fig. 9). The NE555 timer provides a low frequency square wave that is used to supply power to the Vdd pin of the MM5369N. Every time the square wave goes positive, the oscillator is gated on for the duration of the positive portion of the pulse. The oscillator will produce whatever frequency has been selected for it by its crystal.

Construction

These circuits can be wired either point-to-point or by printed circuits. Although my circuit debugging is started on perfboard or breadboarding strips, I have always found it more enjoyable, in the long run, to make all my test equipment on printed circuit board. Not only does it make assembly easier, but

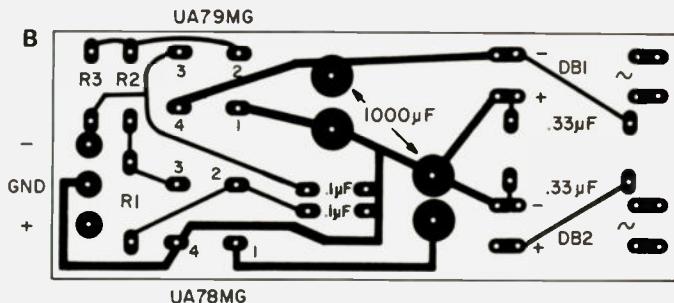
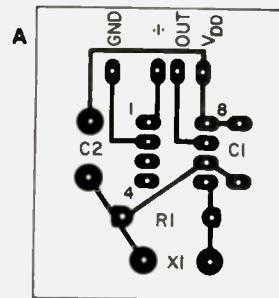


Fig. 10. (a) MM5369N PC board layout for basic circuit and crystal checker. (b) UA78MG and UA79MG power supply PC board layout. Both PC board layouts are shown from the component side.

when properly accomplished, can add to the usefulness and longevity of a circuit. Should you also find this to be the case, PC layouts have been included (Fig. 10).

Calibration

Calibration of the MM5369N output frequency is a relatively simple matter and can be accomplished with a bare minimum of fuss. Hook the output of pin 7 on the MM5369N to an oscilloscope or frequency counter and trim C2 until the desired frequency is obtained. Do this slowly as the trim cap is small and it's easy to overshoot the desired frequency. If you opt for the counter method of calibration and your counter drifts very little, you can record the oscillator output over a period of time while recording the ambient temperature, and make a frequency vs. temperature chart that will allow you to temperature-compensate your oscillator. This is done by finding the rate of frequency drift with temperature and selecting a replacement capacitor for C1 that has the opposite coefficient of the temperature curve.

To select the proper capacitor, note the frequency and temperature at several points. Divide the difference in frequencies (desired and actual) by the desired frequency, then divide the result by the number of degrees that have

changed from your ambient setting and you should have your temperature coefficient in parts per million/degree. This is what capacitor temperature coefficients are expressed in. Simply choose the one that is closest to the opposite of your recorded findings and you will be on your way to maximum stability.

Conclusion

This article is intended to present the experimenter with a versatile, crystal-controlled building block

that proves to be useful in a wide range of circuits and is functional as a general-purpose crystal-controlled frequency standard that is not limited to such a mundane chore as a digital clock timebase. ■

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3. "Brew Up a Signal Generator," 73 Magazine, January, 1978.

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sonal use, the new control box would have to: use LED displays, be relatively simple, interface directly to the rotator with no internal modifications, remain accurate during ac line variations, and match my other station equipment.

Research

Fig. 1 shows the schematic of the TR-44's unmodified sensing circuit. This circuit is a simple voltage divider, which causes more or less current to flow through the meter as the sensing potentiometer (inside the rotator) changes position. When the potentiometer is at the bottom end of its range (rotator fully counterclockwise, as viewed from the top of the rotator), no current flows through the meter, and the meter needle remains in the resting position. When the rotator is at the top end of its range (rotator fully

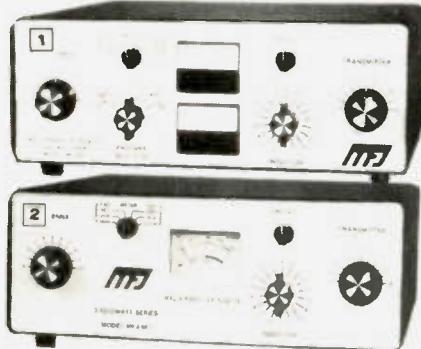


Photograph of the new control box. (The display was enhanced so that it would stand out better.)

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clockwise), approximately one milliamperes flows through the meter, and the needle indicates full scale.

With the above ideas in mind, I decided that a simple dc voltmeter, with proper voltage scaling, could be used to indicate the position of the rotator potentiometer. If you apply a voltage to the potentiometer so the wiper-voltage changes from 0 to 3.6 volts during a complete

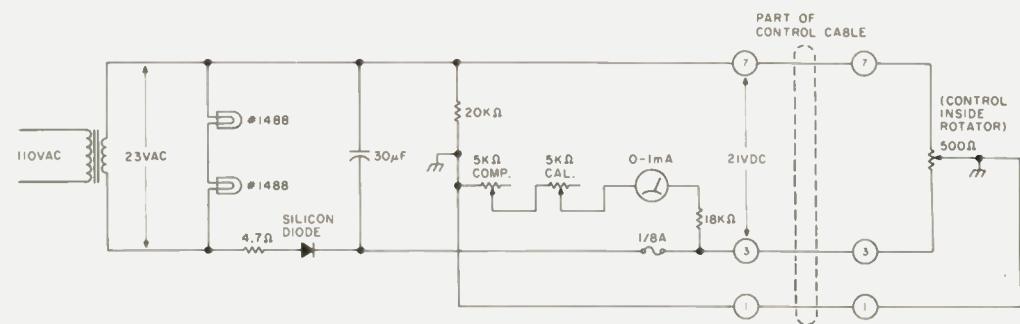


Fig. 1. Unmodified sensing circuit for the TR-44 rotator.

revolution, you have a voltage which changes at the

rate of .01 volts per degree. A preliminary data sheet

for the Motorola MC14433 integrated circuit, which is

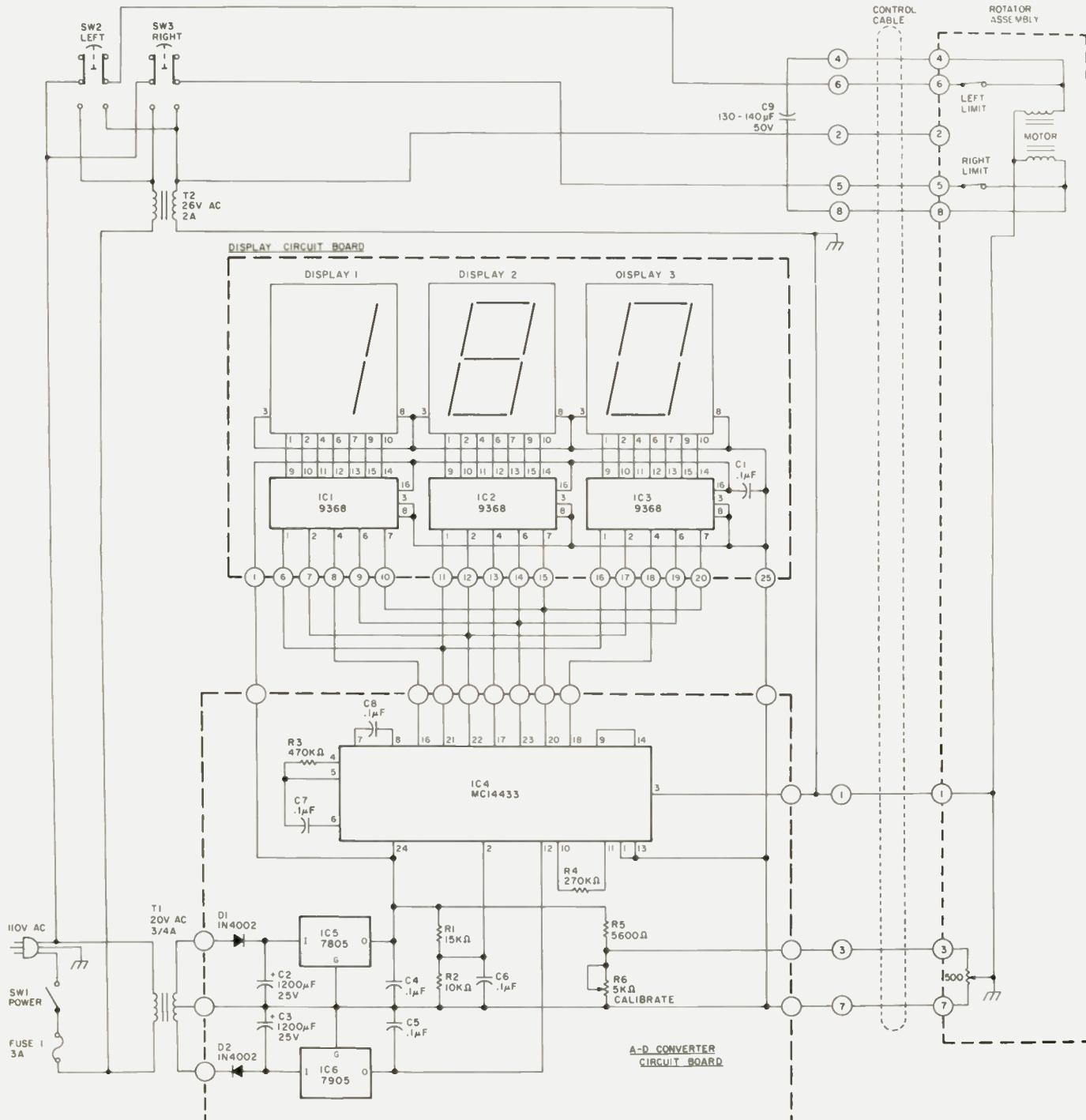


Fig. 2. Schematic diagram of the new control box (shown connected to the original rotator assembly).

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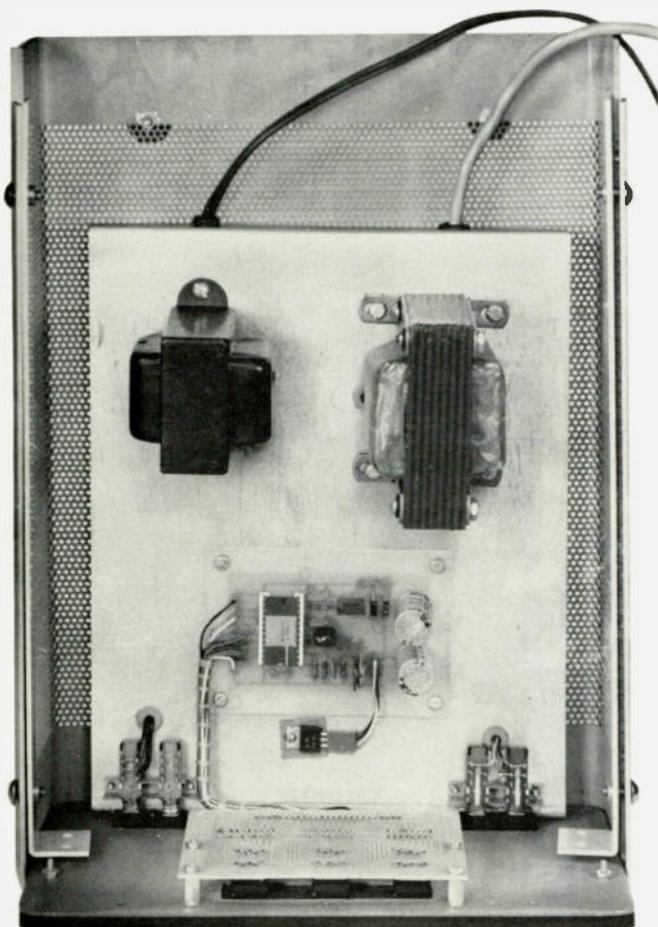
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a 3½-digit analog-to-digital converter, contained the schematic of a simple 3½-digit digital voltmeter. To suit my particular requirements, I deleted the overrange and polarity sensing circuitry to further simplify the circuit. I then

modified the circuit to work with some decoder-driver ICs and LED displays that I already had. Fig. 2 is the schematic diagram of the completely redesigned rotor control box (shown connected to the rotator assembly).



Internal layout of the new control box.

Construction

With one minor exception, construction of the control box is not critical. Since the potentiometer wiper inside the rotator is mechanically grounded to the case (control cable wire #1), you must take care to keep all other circuit points isolated from ground.

If you use the same +5-volt regulator as I did, heat-sink it to the chassis. Be sure to use an insulator between the IC and the chassis to keep it electrically isolated from the chassis. The -5-volt regulator IC does not need a heat sink and may be mounted on the circuit board.

Figs. 3, 4, 5, and 6 show the foil and component layout of the two circuit

boards. Although the two circuit boards could have been combined into one, I used separate boards so that they could be used on future projects.

Be sure the power transformer you use in the sensing circuit has at least an 18-volt, center-tapped, secondary winding. This will allow the regulator ICs to remain within acceptable operating parameters with varying ac line voltages. With a 20-volt, center-tapped transformer, you should be able to keep an accurate display indication while the ac line voltage varies between 90 and at least 130 volts.

I mounted an extra push-button switch on the left side of the front panel to help balance the appearance of the panel. You

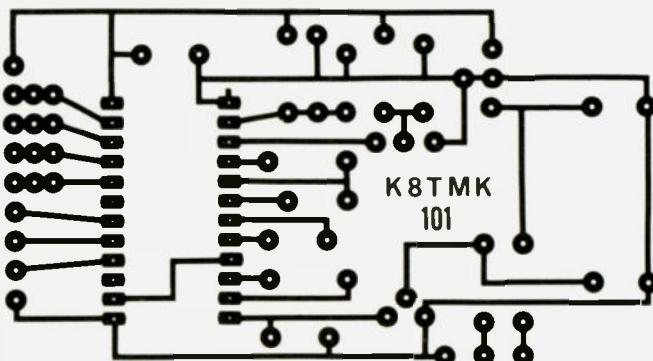


Fig. 3. Actual size foil pattern for the A-D converter circuit board (shown from the foil side).

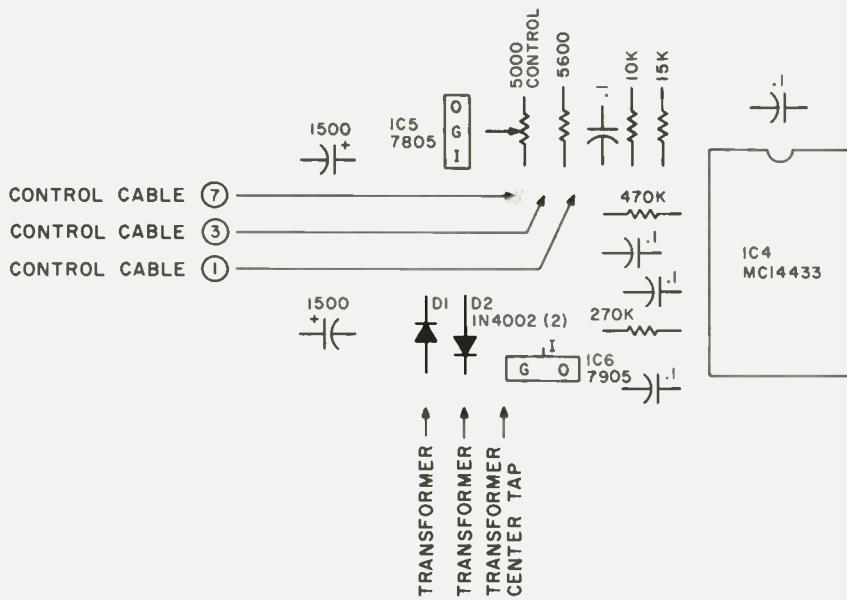


Fig. 4. Component layout for the A-D converter circuit board (shown from the component side).

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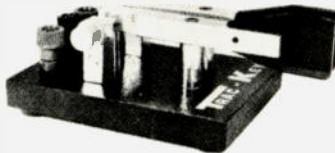
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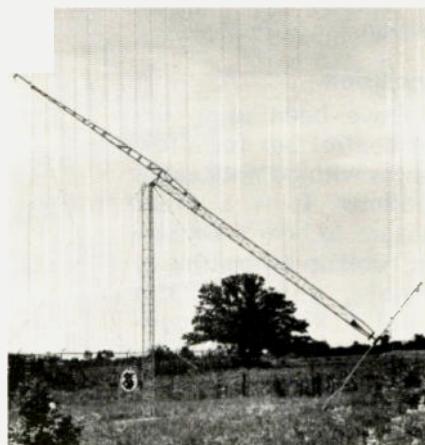
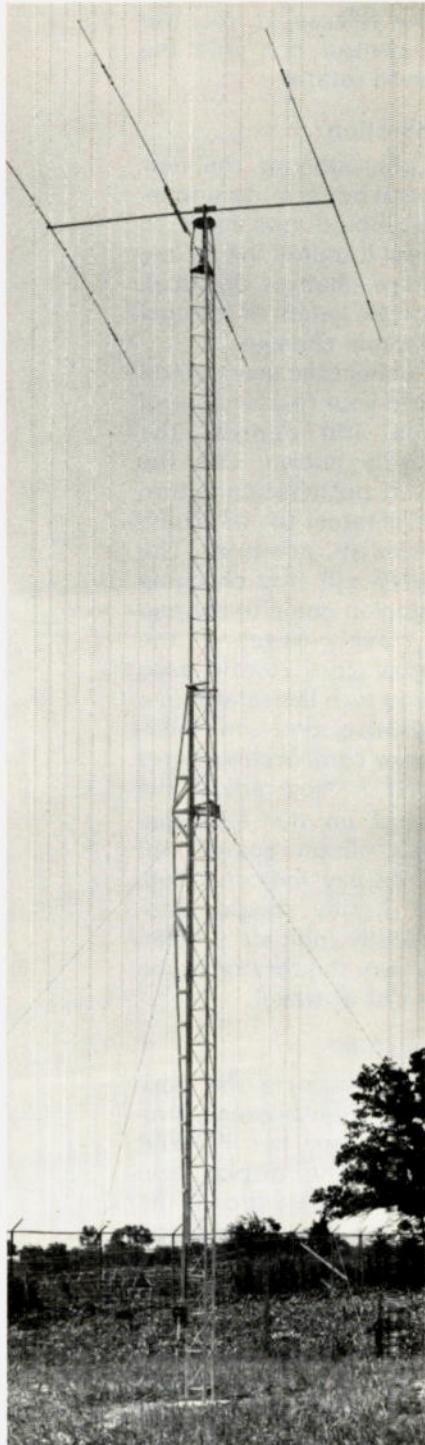
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could use this switch for brake release, if you use the control box with the Ham-M rotator.

Calibration

Calibration of the new control box is quite simple. You should not have to repeat it unless the ac line voltage changes drastically or the length of the control cable changes.

Connect the new control box to your rotator control cable and depress the POWER button. Use the RIGHT push-button to turn the rotator to its fully-clockwise position. The display will stop changing when you come to the end of travel. Note: If the display goes toward zero as you turn the rotator in a clockwise direction, interchange control cable wires 3 and 7. Now adjust the control on the A-D converter circuit board until the display indicates 360. The display should automatically indicate 0 when you turn the rotator to the left end of travel.

Operation

To determine the position of your antenna, simply depress the POWER button. The display will indicate the position in degrees. If you desire, you can leave the POWER button depressed for a continuous indication.

If you wish to turn the rotator to a new position, simply push the LEFT or RIGHT button until the display indicates the desired position.

Conclusion

I have been using the new control box for a few months with no noticeable problems. It is a great pleasure to hear a DX station, look up his bearing in a table, and turn the rotator to the exact position without having to interpolate between the lines on an analog meter.

In the future, I plan to

Component Number	Parts List Description
C1, C4, C5, C6, C8	.1 uF disc ceramic capacitors
C2, C3	1200 uF, 25-V electrolytic capacitors
C7	.1 uF mylar™ capacitor
C9	130-140 uF, 50 V, non-polarized capacitor
D1, D2	1N4002 silicon diode
DISP 1, DISP 2, DISP 3	FND-500 LED displays
F1	3-Ampere, 3AG fuse
IC1, IC2, IC3	9368 7-segment decoder-driver-latch ICs
IC4	MC14433 3½ digit A-D converter IC
R1	15k Ohm, ¼-Watt, 5% resistor
R2	10k Ohm, ¼-Watt, 5% resistor
R3	470k Ohm, ¼-Watt, 5% resistor
R4	270k Ohm, ¼-Watt, 5% resistor
R5	5600 Ohm, ¼-Watt, 5% resistor
R6	5k Ohm linear control
SW1	DPDT locking push-button switch
SW2, SW3	DPDT momentary-contact push-button switch
T1	Power transformer (20 V ac center-tapped, ¾-Ampere secondary)
T2	Power transformer (26 V ac, 2-Ampere secondary)

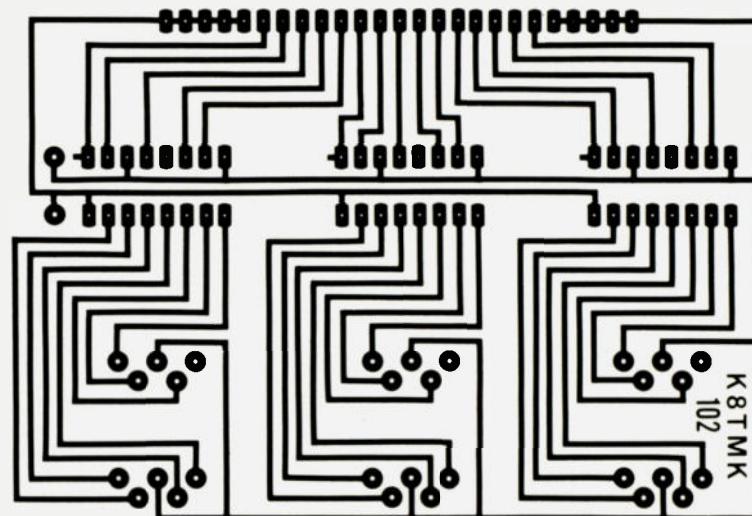


Fig. 5. Actual size foil pattern for the display circuit board (shown from the foil side).

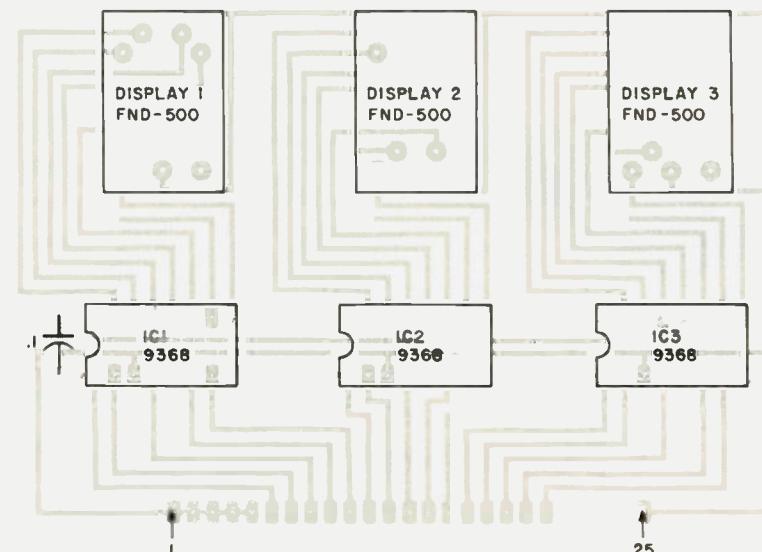


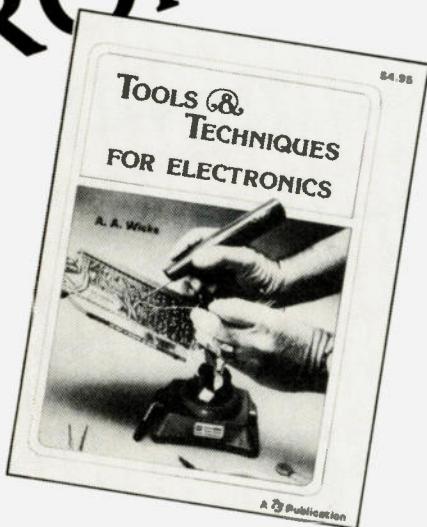
Fig. 6. Component layout for the display circuit board (shown from the component side).

redesign the control box for use with a microprocessor. This will allow me

to punch in a prefix on a keyboard, for example, and then simply wait for

the rotator to reach the correct position automatically. ■

NEW FROM '73



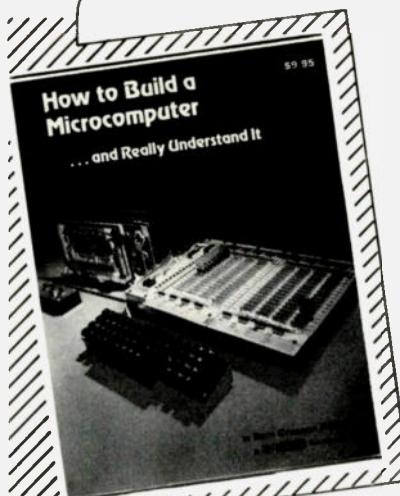
Tools and Techniques for Electronics (BK7348) is a comprehensive guide to the tools and construction practices used by today's electronics hobbyist. This new 73 Magazine publication should be a part of the library of anyone who has ever built or fixed any electronic gear. The text and numerous pictures and illustrations provide an easy-to-understand description of the safe and correct way to use the basic and specialized tools needed for electronics work.

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The short application program listed in Table 1 allows the KIM-1 to send any of three messages by pressing one of three keys (A, B, or C) on the KIM-1 keyboard, and, with the interface circuit shown in Fig. 1, the KIM-1 becomes an electronic keyer as well. Any microcomputer with a

650X microprocessor and one of the MOS Technology PIA or VIA chips may be used with only minor modifications to the program. An important feature of the program is the ability to precisely set the code speed between 5 and 99 words per minute by entering the speed, in decimal, at storage location 0000 in memory. The program converts this decimal number to hexadecimal, then does a division routine to convert the speed to a time duration of the basic dot element, and the interval timers on the 6530 PIA do the rest.

Anyone who does much contest operating will realize how useful an automatic message sender is. Even the casual CW operator can use it for sending CQ or other routine messages. Code tests for Novices can be programmed and sent at precisely 5 wpm by storing the entire test in memory. At 5 wpm, at least 5 minutes of code may be sent. For Field Day (1977), we used a similar program to send CQ CQ CQ FD DE KØEI KØEI K as message A, then when a station responded, we sent () DE KØEI UR 599 MO 599 MO K, where the blank was the call of the station to be keyed by the operator, after which he hit key B to give the remainder of the message. It worked very smoothly with no discernible pause between the call letters and the message. (Don't try to look up our score, because KØEI was not the call we used.) The operation of the keyer is exactly like most electronic keyers; holding the paddle in the dot position will cause a series of dots and spaces to be sent. Dashes occur with the paddle in the dash position, and the timing of all the characters is controlled by the program and the crys-

tal on the microcomputer.

Operation

Assuming the program has been loaded and the interface circuit connected, operation proceeds as follows. The code speed at which you wish to operate is loaded into storage location 0000. Any decimal number from 05 to 99 may be put into this location. Next, the starting and ending addresses of each message must be loaded into memory. Since all three messages are in page 2 of memory, only the low-order bytes of the starting and ending addresses need be given. Suppose message A starts at 0200 and ends at 0251, message B starts at 0252 and ends at 0265, while message C starts at 0266 and ends at 0278. (These are completely weird numbers chosen at random from my skull and have absolutely no other significance.) Then one would load 00, the starting address of message A, at location 0001; 52, the starting address of message B goes in at 0002; and 66 is entered at 0003. The respective ending addresses go into memory locations 0004 to 0006, that is, 51 into 0004, 65 into 0005, and 78 into 0006.

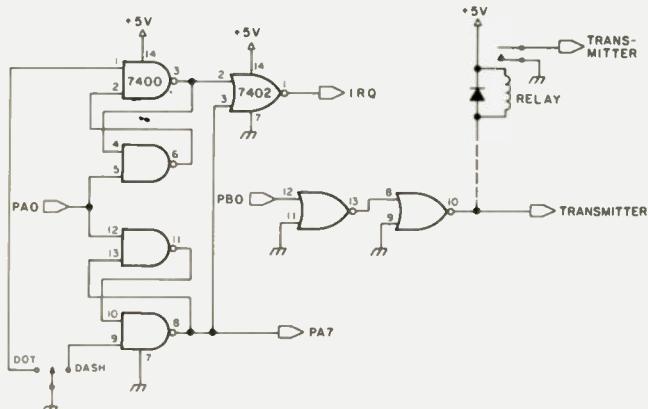


Fig. 1. Interface circuit. Some transmitters may require the optional relay for keying, with a 1N914 diode across the coil for protection against transient voltages. The Triton IV may be keyed directly from pin 10 of the 7402. All grounds should be the same as the KIM-1 ground, and the 5 volts may be stolen from the KIM-1 power supply.

How do you load the messages themselves? For each character you want to send, you must load the corresponding hex number shown in Table 2. Suppose, for no reason whatsoever, message A is to be "DE KØEI K" and is to start at 0200. You then load the hex numbers 90 40 00 B0 FC 40 20 00 B0 from locations 0200 through 0208; 00 goes into 0001 and 08 into 0004.

Probably the best way to proceed is to first load the three messages including spaces, noting the starting and ending addresses of each message on a piece of paper. Then go back to page 0 and put the starting and ending addresses in their proper locations. Go to location 0300 and hit the GO button to start the program running. Test to make sure everything is working before you put the thing on the air.

Programming Details

The flowchart shown in Fig. 2 and the comments in the program should give the reader a good feeling

for the structure of the program. It consists of three principle parts: the main program, subroutine SEND, and the interrupt routine, all of which have individual flowcharts shown. Minor components are subroutine DIT (which holds PB0 at logic 0 for the dot length followed by a logic 1 for the space length), subroutine DAH (which holds PB0 at logic 0 for three dot lengths—1 dah = 3 dits—followed by a space), and subroutine TIMER (which loads the timer on the KIM-1 with the precise length of 1 dot and then waits for this time to elapse).

We now look at some specific details of the program. The speed in words per minute must be converted to hex before the computer can do any further calculations with it. This conversion may best be explained with an example. Suppose we wish to operate at 20 wpm, so 20 is entered into location 0000. What we mean by 20 is 2 in the tens place and 0 in the

ones place, but what the computer thinks this means is 2 in the sixteens place and 0 in the ones place. At least we agree on the ones place, so initially we mask the ones place out with an AND statement; later we retrieve it and simply add it to the result of our decimal-to-hex conversion of the 2. To trick the computer into thinking the 2 in the sixteens place is the 2 in the tens place we intended it to be, we change the sixteen to a ten with this trick: $10 = 16/2 + 16/8$. The sixteens place divided by two is accomplished by one shift-right statement (LSR), while the sixteens place divided by 8 is accomplished by three shift-right statements. So, the 2 in the sixteens place is shifted right once, stored, shifted right two more times, and these two results are added. We now have 2×10 in the computer (in hex, of course) rather than 2×16 . Adding the results from the ones place completes the conversion.

Using the keying speed definitions from *The Radio Amateur's Handbook*, one can calculate that the dot length in milliseconds is $1200/S$, where S is the code speed in words per minute. If the divide-by-1024 timer on the KIM-1 is used, 1 count corresponds to 1.024 milliseconds. Converting the dot length to timer counts gives $TIME = (1172/S)$ base 10 = $(494/S)$ hex, where TIME is the number to be loaded into the divide-by-1024 timer to give a code speed of S wpm. So, the computer must divide S into 494. This is determined by successively subtracting S from 494 until the result becomes negative. The number of subtractions is the quotient of $494/S$.

Pin PB0 on the KIM-1 is used as the keying output from the computer. When power is applied to the computer and the reset button is depressed, PB0 comes up in a logic 1 state. This dictates that logic 1 corresponds to the transmitter being off. Consequently, PB0 is buffered and inverted twice by the NOR gates. Inverters such as the 7404 would work, but since I needed a NOR gate in the keyer interface, I simply used the other NOR gates on the same

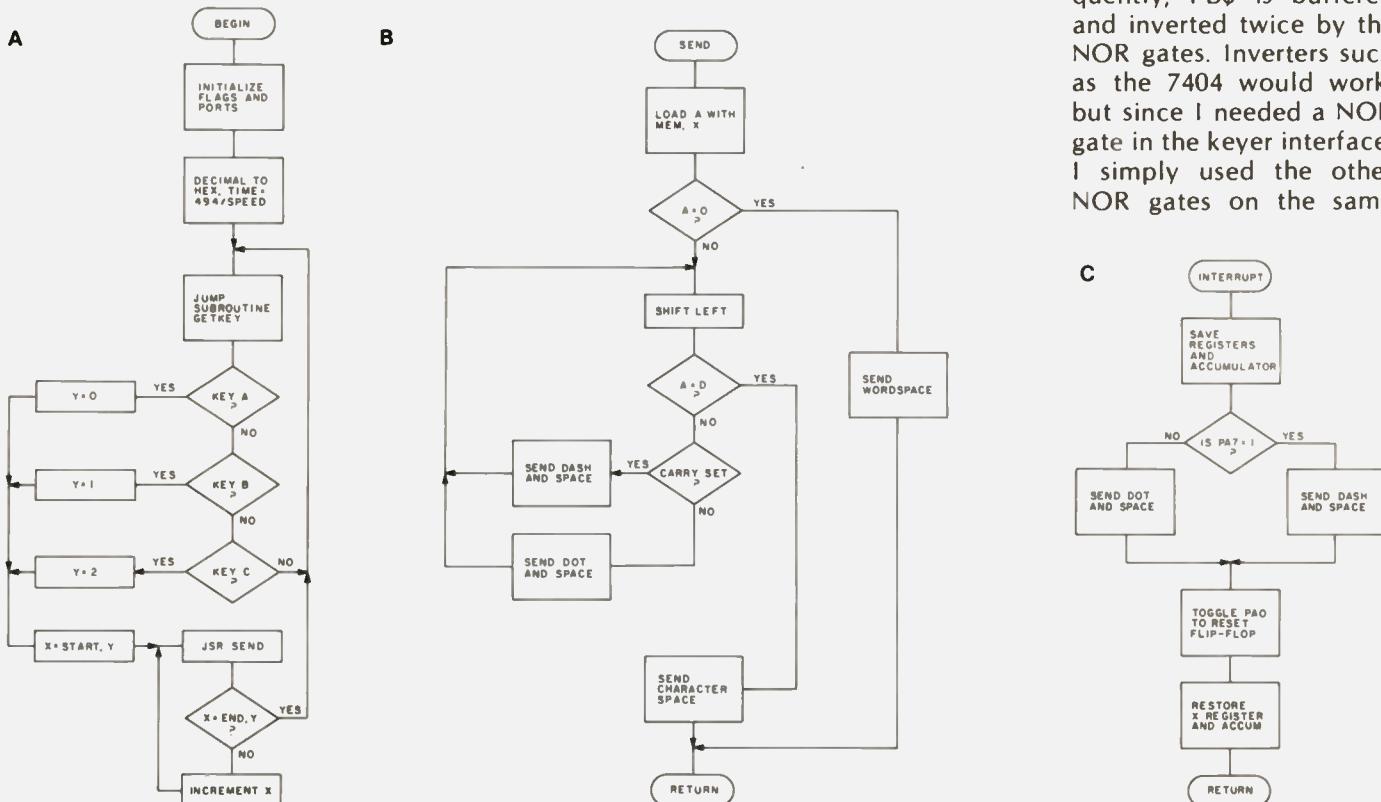


Fig. 2. Flowchart for the keyer and message sender. (a) Main program. (b) SEND subroutine. (c) Interrupt routine.

Address	Instruction	Label	Op Code	Operand	Comments	0376	8A	SEND	TXA
0300	78	BEGIN	SEI		Prevent interrupts.	0377	48	PHA	
0301	D8		CLD		Binary mode.	0378	BD 00 02	LDA MEM,X	Get code element.
0302	A9 C9		LDA \$C9		Set interrupt vectors.	037B	F0 1E	BEQ WDSP	
0304	8D FE 17		STA IRQL			037D	QA	HERE	ASL A
0307	A9 03		LDA \$03			037E	F0 10		BEQ FINSH
0309	8D FF 17		STA IRQH			0380	48	PHA	
030C	A9 01		LDA \$01		Initialize I/O Ports A	0381	B0 06	BCS DASH	
030E	8D Q2 17		STA PBD		and B.	0383	20 AD 03	JSR DIT	Send dot.
0311	8D 03 17		STA PBDD		PB0 is output pin.	0386	4C 8C 03	JMP ARND	
0314	8D 01 17		STA PADD		PA0 is output pin.	0389	20 B9 03	DASH	JSR DAH
0317	8D 00 17		STA PAD			038C	68	ARND	PLA
031A	CE 00 17		DEC PAD		Toggle PA0 to reset debounce	038D	4C 7D 03		JMP HERE
031D	EE 00 17		INC PAD		circuit.	0390	A2 02	FINSH	LDX \$02
0320	A5 00		LDA SPEED		Get decimal value of speed	0392	20 BE 03	AGN	JSR TIMER
0322	48		PHA		from location 0000 and convert	0395	CA		Character space.
0323	29 F0		AND \$FO		it to hex.	0396	DO FA		
0325	4A		LSR A		Multiply tens digit by ten.	0398	68		
0326	85 10		STA SCRATCH			0399	AA		
0328	4A		LSR A			039A	60		
0329	4A		LSR A			039B	A2 04	WDSP	LDX \$04
032A	18		CLC			039D	4C 92 03		JMP AGN
032B	65 10		ADC SCRATCH						*****
032D	85 10		STA SCRATCH		Result of multiplication here.	03A0	A2 01	DIT	LDX \$01
032F	68		PLA		Get SPEED again.	03A2	CE 02 17	BACK	DEC PBD
0330	20 0F		AND \$0F		Add ones digit to SCRATCH.	03A5	20 BE 03	SPA	JSR TIMER
0332	65 10		ADC SCRATCH			03A8	CA		
0334	85 10		STA SCRATCH		Decimal to hex complete.	03A9	DO FA		
0336	38		SEC		Division routine begins here.	03AB	AD 02 17		
0337	A2 00		LDX \$00			03AE	4A		
0339	A9 94		LDA \$94			03AF	B0 07		
033B	85 08		STA LO			03B1	EE 02 17		
033D	A9 04		LDA \$04			03B4	E8		
033F	85 09		STA HI			03B5	4C A5 03		
0341	A5 08	UP	LDA LO			03B8	60	DONE	RTS
0343	E5 10		SBC SCRATCH						*****
0345	85 08		STA LO			03B9	A2 03	DAH	LDX \$03
0347	A5 09		LDA HI			03BB	4C A2 03		JMP BACK
0349	E9 00		SBC \$00						*****
034B	85 09		STA HI			03BE	A5 07	TIMER	LDA TIME
034D	E8		INX			03C0	8D 07 17		Delay for the number of
034E	B0 F1		BCS UP			03C3	2C 07 17	CHK	STA TIMER
0350	86 07		STX TIME		Division Complete.	03C6	10 FB		1.024 millisecond units
0352	20 6A 1F	RPT	JSR GETKEY		Read keyboard subroutine.	03C8	60		stored in TIME.
0355	58		CLI			03C9	48	INTERRUPT	Save registers.
0356	A0 00		LDX \$00		Test keys.	03CA	8A		
0358	C9 QA		CMP \$QA			03CB	48	PHA	
035A	F0 QA		BEQ MESSA			03CC	AD 00 17	LDA PAD	
035C	C9 QB		CMP \$QB			03D0	30 06	BMI PAST	Is PA7 = logic 1?
035E	F0 05		BEQ MESSB			03D2	20 AO 03	JSR DIT	Yes, dah. No, dit.
0360	C9 OC		CMP \$OC			03D5	4C DA 03	JMP ACRS	Send dot.
0362	D9 EE		BNE RPT			03D8	20 B9 03	PAST	JSR DAH
0364	C8		INY			03DB	CE 00 17	ACRS	Send dash.
0365	C8	MESSB	INY			03DD	EE 00 17	DEC PAD	
0366	BE 01 00	MESSA	LDX STRT,Y		Start message.	03E0	68	INC PAD	Toggle debounce circuit.
0369	20 76 03	CNT	JSR SEND			03E1	AA		
036C	8A		TXA			03E2	68	PLA	Restore registers.
036D	D9 04 00		CMP END,Y		End message?	03E3	40		
0370	F0 E0		BEQ RPT						Return from interrupt.
0372	E8		INX						
0373	4C 69 03		JMP CNT						

Table 1. Source listing for the message and keyer program.

chip. If PB0 could sink enough current, it might drive the relay directly, but I preferred the buffering shown in Fig. 1. Mark elements of the Morse code are sent by decrementing (DEC) PB0 for the appropriate length of time,

while space elements are sent by leaving PB0 at logic 1.

The program idles in the loop starting with JSR GETKEY and ending with BNE RPT, testing each of three keys (A, B, and C) to see if they were depressed.

If no key is depressed, the program remains in this loop. If a key is depressed, register Y is set to 0, 1, or 2 depending on which key was struck. Y is then used as an index to look up the starting address (low-order byte of page two of memory) of the message (STRT,Y) and later the ending address (END,Y) of the message. The starting address is used as an index to find the first code element of the message (MEM,X), and it is incremented until the ending address is en-



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A	60	X	98	70
B	88	Y	B8	0001
C	A8	Z	C8	0002
D	90	1	7C	0003
E	40	2	3C	0004
F	28	3	1C	0005
G	D0	4	0C	0006
H	08	5	04	00F1
I	20	6	84	
J	78	7	C4	
K	B0	8	E4	
L	48	9	F4	
M	E0	0	FC	
N	A0	Word space	00	
O	F0	SK	16	
P	68	BT	8C	
Q	D8	AR	54	
R	50	/	94	
S	10	.	56	
T	C0	,	CE	
U	30	?	32	

Table 2. Morse character to hex conversion.

countered.

The conversion of an 8-bit word of memory to a Morse code character has been described in other references in detail and will not be repeated here. There are a number of

schemes available,^{1,2} but the most efficient schemes appear to be those in references 3 and 4, and those were the techniques used here.

The keyer is implemented by the interrupt routine,

Location	Contents
0000	Speed in decimal (words per minute)
0001	Starting address of message A (low-order byte)
0002	Starting address of message B
0003	Starting address of message C
0004	Ending address of message A (low-order byte)
0005	Ending address of message B
0006	Ending address of message C
00F1	04 (Prevents interrupts while in monitor)

Table 3. Storage locations to be loaded by the operator.

which in turn uses subroutines DIT, DAH, and TIMER. It will send at exactly the same speed as the messages. The keyer interface circuit is simply debouncers which are reset at the end of an interrupt. If the key is still in the dot or dash position, the reset has no effect and another interrupt occurs. The flowchart indicates that the state of PA7 determines which element is to be sent.

One last thought: If you want to be able to key in a few characters in the middle of a message, just load a few word spaces there and key the characters in

when the blank occurs. This is handy for giving signal reports and also in some contests where the number of contacts is updated after each QSO. ■

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1. Pollock, James W., "1000 WPM Morse Code Typer," 73, January, 1977, p. 100.
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3. Pollock, James W., "A Microprocessor Controlled CW Keyboard," Ham Radio, January, 1978, p. 81.
4. Ockers, Stan, "Code Test," The First Book of KIM, ORB, Argonne, Illinois, 1977, p. 56.

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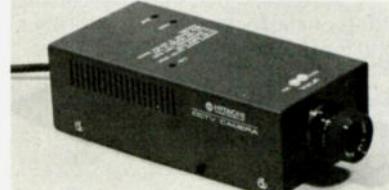
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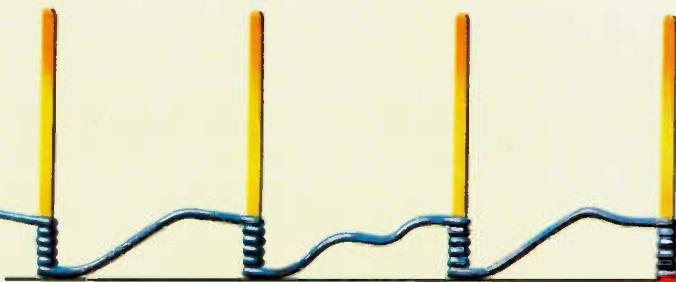
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MORTGAGE WITH PREPAYMENT OPTION/FINANCER

These two programs will more than pay for themselves if you mortgage a home or make investments:

• **Mortgage with Prepayment Option**—Calculate mortgage payment schedules and save money with prepayments.

• **Financier**—Calculate which investment will pay you the most, figure annual depreciation, and compute the cost of borrowing, easily and quickly.

All you need to become a financial wizard with an 8K PET. Order No. 0006P

Casino I 8K; Order No. 0014P

Casino II 8K, Order No. 0015P.

Checkers/Baccarat 8K Order No. 0022P.

INSTANT SOFTWARE

Dow Jones

*A trademark of Commodore Business Machines Inc.

CLOSING PRICE QUOTATIONS DAY 6			
INDUSTRIAL IN PR4	CUB PR4	CHG4-A	
AT&T STEEL 44	44.250	+1.75	
IBM MRS 31	14.500	-1.50	
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AT&T VEH 21	21.317	-1.562	
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DOW JONES Up to six players can enjoy this exciting stock market game. You can buy and sell stock in response to changing market conditions. Get a taste of what playing the market is all about. Requires a PET with 8K. Order No. 0028P

Tangle/Supertrap 8K; Order No. 0029P

Trek-X 8K, Order No. 0032P.

Qubic-4/Go-Moku 8K, Order No. 0038P

MIMIC

How good is your memory? You can find out with the Mimic program, and for a fraction of the price of similar games being offered on the market today.

Your PET will flash a sequence of figures on the screen and ask you to repeat them in the same order. It may sound easy, but there are five different programs, one of which is sure to keep you going until you're ready to throw in the towel. It's fun, it's bound to improve your memory, and it's even programmed for optional sound effects; all you need is an amplifier and a speaker.

Order No. 0039P

Penny Arcade 8K, Order No. 0044P

Arcade II 8K, Order No. 0045P

Business Manager 8K, Order No. 0062P \$14.95

Dungeon of Death 8K, Order No. 0064P

Arcade I 8K, Order No. 0074P

Digital Clock 8K, Order No. 0083P

Turf and Target 8K, Order No. 0097P

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**A trademark of Commodore Business Machines Inc.

***A trademark of Apple Computer Inc.

Now you can get the best for less! Just clip out the coupon below, bring it to your local computer store, and you'll get \$1.00 off the list price of any program package from Instant Software!

Apple ★★★

Golf Applesoft II + 20K Order No. 0018A.
Bowling/Triology 20K: Order No. 0040A

MATH TUTOR I The Math Tutor I package is designed to help younger children learn the concepts of mathematics. It uses the technique of positive, immediate reinforcement. When the child enters a correct answer, he or she is rewarded with the opportunity to move a spaceship, cast a magic spell, or rescue a man from a most unfortunate fate. By using this method of positive reinforcement, children will learn to associate mathematics

with pleasant experiences, as opposed to the stark reality of numbers that has driven so many pupils to an early dislike of math. The program also provides the opportunity for parents to indulge in a unique learning experience with their children, while introducing them to the exciting world of microcomputing. Teachers can also use the program in a classroom setting to provide that extra incentive. Math Tutor I contains the following programs:

- **Hanging**—A variation of the traditional game of Hangman. In Hanging, students must spare the graphic character's "life" by correctly answering the questions posed by the computer.
- **Spellbinder**—The student is cast as a magician competing against another magician who's controlled by the computer. He must answer the questions about fractions.

tions correctly in order to cast magic clouds, fireballs, and other spells at his rival.

- **Whole Space**—Students are given the opportunity to lead an attack on an evil planet. Every time a correct answer is given to a math problem posed by the computer, he can move his ship closer to his goal. Every wrong answer gives the enemy an opportunity to fire back.

Note: the Hanging program requires Applesoft II BASIC. The Whole Space and Spellbinder programs can run in integer BASIC. Order No. 0073A.

Math Tutor II Applesoft II ± 20K Order No. 0098A

Dealer Inquiries Invited

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Make a great buy even better . . .

Why pay full price? Take this coupon to the store nearest you and get \$1.00 off on any of our software. If the store nearest you doesn't stock Instant Software, don't call the manager names—there's still hope. Just tell him which programs you want, and he'll order the software for you. If you're so wealthy that saving a buck means nothing to you—not to mention the added dollar handling charge—then use the order blank at right to order your software direct, or call Toll-Free 1-800-258-5473.



DSI HAS DONE IT AGAIN

QUIK-KIT II®

INCLUDES PROPORTIONAL OVEN TIME BASE

- DC-BATT-AC (W-AC9)
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- 100% Factory Tested
- .2 PPM 10° to 40° C Accuracy
- 9 Digits .5 Inch LED's
- 0.1 HZ Resolution
- Auto Zero Blanking

\$149.95

MODEL 5600A KIT



WHY BUY A 5600A: Because 95% of the assembly is completed by DSI and you are only one hour away from solving all those difficult bench problems, from setting the frequency of a audio signal to within 1/10 of a HZ, to checking the frequency of a 486 MHZ mobile radio. Whether you are servicing a VTR, trouble shooting a PLL circuit, the 5600A is the right counter with accuracy that will meet any FCC land mobile, broadcast, or telecommunications requirements. On the bench or in the field the 5600A will do the job you need. The 5600A includes a self contained battery holder providing instant portability or we offer a 10 hour rechargeable battery pack option. Other options include a audio multiplier which allows you to resolve a 1/1000 of a HZ signal and finally a 25db preamplifier with an adjustable attenuator making the 5600A perfect for communications, TV servicing, industrial testing or meeting your QSO on the correct frequency every time.

FACTS ARE FACTS: With the introduction of the 5600A. The sun has set on the competition. This may sound like a bold statement on the part of DSI BUT FACTS ARE FACTS. No counter manufacturer except DSI offers a Full Range 50 HZ to 600 MHZ counter with — 9 Digits — 0.1 HZ resolution — .2 PPM 10° to 40 ° C proportional oven — RF pre-amp — 600 MHZ prescaler — three selectable gate times — oven ready, standby and gate time indicator lights as standard features — For only \$149.95 kit and \$179.95 factory wired. In fact the competition doesn't even come close unless you consider \$200.00 to \$800.00 close. With DSI having the best price to quality features ratio in the industry, no wonder we've become one of the world's largest manufacturers of high quality frequency counter instrumentation.

**FOR INFORMATION — DEALER LOCATION — ORDERS — OEM
CALL 800-854-2049 CALIFORNIA RESIDENTS CALL 800-542-6253**

Model	Price	Frequency Range	Accuracy Over Temperature	Sensitivity			Number of Readouts	Size of Readouts	Power Requirements	H W D
				@ 100Hz-25MHz	@ 50-250MHz	@ 250-450MHz				
5600A-K	\$149.95	50Hz-600MHz	Proportional Oven .2 PPM 10° - 40° C	10MV	10MV	50MV	9	.5 Inch	115 VAC or 8.2-14.5 VDC	3 1/4" x 9 1/2" x 9"
5600A-W	\$179.95									
3550	99.95	50Hz-550MHz	TCXO 1 PPM 17° - 40° C	25MV	25MV	75MV	8	.5 Inch	115 VAC or 8.2-14.5 VDC	2 1/8" x 8" x 5"
500HH	\$149.95	50Hz-550MHz	TCXO 1 PPM 17° - 40° C	25MV	20MV	75MV	8	.4 Inch	115 VAC or 8.2-14.5 VDC or NICAD PAK.	1" x 3 1/2" x 5 1/2"

5600A wired factory burned in 1 year limited warranty 5600A kit 90 day limited warranty
Prices and/or specifications subject to change without notice or obligation

*With AC-9 Adaptor

\$99.95



MODEL 3550K

3550 OWNERS
You can add the
35P.2 .22 PPM
10° to 40° C
proportional oven
to your
existing 3550



DSI INSTRUMENTS, INC.
7924 Ronson Road
San Diego, California 92111

T101 Ant.....	\$3.95
AC-9 AC Adaptor.....	7.95
35P.2	29.95
Factory Installed	49.95

TERMS: MC - VISA - AE - Check - M.O. - COD in U.S. Funds
Please add 10% to a maximum of \$10.00 for shipping, handling and insurance. Orders outside of USA & Canada, please add \$20.00 addition to cover air shipment. California residents add 6% Sales Tax.

5600A Kit	\$149.95
5600A Wired	179.95
AC-9 AC Adaptor.....	7.95
T600 BNC Ant.....	7.95

BUILT-IN OPTIONS

BA56 Rechargeable 10 Hr. Bat. Pack	24.95
AM56 Audio Multiplier .001Hz Resolution	34.95
PA56 25dB Preamplifier with Attenuator	59.95

NEW FROM DSI!!

50 Hz — 500 MHz 8 DIGITS
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- AC-DC Operation
- BNC Inputs 1 Meg Direct 50 Ohms Prescaled
- 8 Large .4" LED Readouts
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\$149⁹⁵

MODEL 500 HH
 50 Hz — 500 MHz
 Without Battery Capability

SAVE \$500

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MODEL 500 HH .. \$169.95
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 Includes AC-9 Battery Eliminator

MODEL 100 HH
 50 Hz — 100 MHz

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Without Battery Capability

The 100 HH and 500 HH hand held frequency counters represent a significant new advancement, utilizing the latest LSI design . . . and because it's a DSI innovation, you know it obsoletes any competitive makes, both in price and performance. No longer do you have to sacrifice accuracy, ultra small readouts and poor resolution to get a calculator size instrument. Both the 100 HH and 500 HH have eight .4 inch LED digits — 1 Hz resolution — direct in only 1 sec. or 10 Hz in .1 sec. — 1 PPM TCXO time base. These counters are perfect for all applications be it mobile, hilltop, marine or bench work. **CALL TODAY TOLL FREE:** (800—854-2049) **Cal Res. CALL** (800—542-6253) **TO ORDER OR RECEIVE MORE INFORMATION ON DSI'S FULL PRODUCT LINE OF FREQUENCY COUNTERS RANGING FROM 10 Hz TO 1.3 GHz.**

FREQUENCY COUNTER CONSUMER DATA COMPARISON CHART

MANUFACTURER	MODEL	SUG'STD. LIST PRICE	FREQUENCY RANGE	TYPE OF TIME BASE	ACCURACY OVER TEMPERATURE		SENSITIVITY			DIGITS		PRE-SCALE INPUT RESOLUTION	
					17° - 40°C	0° - 40°C	100 Hz - 25 MHz	50 MHz - 250 MHz	250 MHz - 450 MHz	No.	SIZE IN INCHES	.1 SEC	1 SEC
DSI INSTRUMENTS	100 HH	\$ 99.95	50Hz-100MHz	TCXO	1 PPM	2 PPM	25 mV	NA	NA	8	4	100 Hz	10 Hz
DSI INSTRUMENTS	500 HH	\$149.95	50Hz-550MHz	TCXO	1 PPM	2 PPM	25 mV	20 mV	30 mV	8	4	100 Hz	10 Hz
CSC‡	MAX-550	\$149.95	1kHz-550MHz	Non-Compensated	3 PPM @ 25°C	8 PPM	500 mV*	250 mV	250 mV	6	1	NA	1 kHz
OPTOELECTRONICS	OPT-7000	\$139.95	10Hz-600MHz	TCXO	1.8 PPM	3.2 PPM	NS	NS	NS	7	4	1 kHz	100 Hz

* 1 KHz - 50 MHz ‡ Continental Specialties Corp.

The specifications and prices included in the above chart are as published in manufacturer's literature and advertisements appearing in early 1979. DSI INSTRUMENTS only assumes responsibility for their own specifications.

100 HH... \$ 99.95 W/Battery Pack... \$119.95
 500 HH... \$149.95 W/Battery Pack... \$169.95

Prices and/or specifications subject to change without notice or obligation.

These prices include factory installed rechargeable NiCad battery packs.



DSI INSTRUMENTS, INC.
 7924 Ronson Road, Dept. G
 San Diego, California 92111

T-500 Ant. \$ 7.95
AC-9 Battery Eliminator \$ 7.95

TERMS: MC - VISA - AE - Check - M.O. - COD in U.S. Funds. Please add 10% to a maximum of \$10.00 for shipping, handling and insurance. Orders outside of USA & Canada, please add \$20.00 addition to cover air shipment. California residents add 6% Sales Tax.

CALL TOLL FREE

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Communications Center

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 Lincoln, Nebraska 68504
 In Nebraska Call (402)466-8402

✓ C58



Antenna Sale!

HY-GAIN

		Regular	Special		Regular	Special
TH6DX	Super Thunderbird	\$299.95	\$239.95	18HT	Hy-Tower 80-10M vertical	299.95
TH3MK3	3 el. 10-15-20M beam	229.95	179.95	18AVT/WB	80-10M Trap vertical	99.95
TH3JR	3 el. 10-15-20M beam	149.95	129.95	14AVQ/WB	40-10M Trap vertical	69.95
Hy-Quad	2 el. 10-15-20M Quad	229.95	179.95	12AVQ	20-10M Trap Vertical	39.95
205BA	5 el. "Long John" 20M beam	289.95	229.95	14RMQ	Roof Mounting kit (verticals)	33.95
155BA	5 el. "Long John" 15M beam	169.95	139.95	5BDQ	80-10M Trap doublet	89.95
105BA	5 el. "Long John" 10M beam	119.95	99.95	2BDQ	80-40M Trap doublet	49.95
204BA	4 el. 20M beam	219.95	179.95	66B	6 el. 6M beam	119.95
204MK5	5 el. conversion kit	99.95	79.95	203	3 el. 2M beam	15.95
153BA	3 el. 15M beam	79.95	69.95	205	5 el. 2M beam	17.95
103BA	3 el. 10M beam	54.95	44.95	208	8 el. 2M beam	25.95
402BA	2 el. 40M beam	209.95	169.95	214	14 el. 2M beam	31.95
BN-86	Balun for beam antennas	15.95	15.95	LA-1	Deluxe lightning arrestor	59.95
TH2MK3	2 el. 10-15-20M beam	149.95	119.95			49.95

MOSLEY

	Regular	Special
Classic 33	3 el. 10, 15, 20 Mtr. beam	304.75
Classic 36	6 el. 10, 15, 20 Mtr. beam	392.75
TA-33	3 el. 10, 15, 20 Mtr. beam	264.00
TA-36	6 el. 10, 15, 20 Mtr. beam	392.75
TA-33 Jr.	3 el. 10, 15, 20 Mtr. beam	197.00
TA-40KR	40 Mtr. Add On	119.95
		89.95

CUSHCRAFT

ATB-34	4 ele. 10, 15, 20 Mtr. beam	289.95	219.95	A147-11	11 ele. 146-148 Mhz. beam	36.95	30.95
ATV-4	10, 15, 20, 40 Mtr. Vertical	89.95	69.95	A147-22	22 ele. Power Pack	109.95	89.95
ATV-5	10, 15, 20, 40, 80 Mtr. Vertical	109.95	89.95	A144-10T	2 Mtr. "Twist" 10 ele.	42.95	34.95
ARX-2	2 Mtr. Ringo Ranger	39.95	32.95	A144-20T	2 Mtr. "Twist" 20 ele.	62.95	52.95
AR-6	6 Mtr. Ringo	36.95	32.95	A147-20T	2 Mtr. beam	62.95	52.95
ARX-220	220 Mhz. Ringo Ranger	39.95	32.95	A430-11	432 Mhz. 11 ele. beam	34.95	29.95
ARX-450	435 Mhz. Ringo Ranger	39.95	32.95	A432-20T	430-436 Mhz. Beam	59.95	49.95
A144-11	11 ele. 144-146 Mhz. beam	36.95	30.95				

HUSTLER

3-TBA	3 ele. 10, 15, 20 Mtr. beam	259.95	189.95
4-BTV	10-40 Mtr. Vertical	99.95	79.95
5-BTV	10-80 Mtr. Vertical	134.95	99.95
RM-75	75 Meter Resonator	16.95	14.50
RM-75S	75 Meter Super Resonator	31.95	27.50
G6-144B	2 Mtr. Base Colinear	79.95	59.95
G7-144	2 Mtr. Base Colinear	119.95	89.95

WILSON

System One	5 ele. 10, 15, 20, Mtr. Beam	\$299.95	\$239.95
System Two	4 ele. 10, 15, 20 Mtr. Beam	249.95	199.95
System Three	3 ele. 10, 15, 20 Mtr. Beam	199.95	159.95
WV-1	10-40 Mtr. Vertical	79.95	69.95

TAYLOR

HQ 1040AV	10-40 Mtr.	59.95	49.95
	Trap vertical		

ROTORS

HamIII \$125.00 T2X Tailtwister \$199.95

Call for prices on rotor cable, Coax, Towers, and Accessories.

Alliance HD 73 \$109.95

All prices do not include shipping.

18HT

We carry all major brands of ham radios
 AT DISCOUNT PRICES

Yaesu — Kenwood — Drake — ICOM — Dentron —
 Ten-Tec — Swan — Tempo — Midland — E.T.O. — Wilson



ATTENTION:

TO: All Amateurs
FROM: Wilson Systems, Inc.

Inflation . . . gas shortages . . . etc., all leading to higher prices each week, and cutting us sane in this runaway inflation period, our escape from the hustle and hectic grind of working to make a living. We know — we see the same price increases at the grocery store, the same increases in the gas prices. Wilson Systems, Inc., is going to do something to help ease the purchase of your new tower and antenna.

As you may know, in January of 1979, Regency Electronics, Inc., purchased Wilson Electronics Corp. What you may not know is that in August, 1979, Jim Wilson purchased back the antennas and towers. There is now a new name to look for — WILSON SYSTEMS, INC. — With the new name and new company comes new ideas, methods, products and prices. Yes, prices. But not what you might expect. Wilson Systems is LOWERING the prices to where you will find it hard to believe. Check them out in the following pages of this issue.

You will be surprised and pleased at what you will find.
What are we doing that will enable us to lower the prices? Well, we are Hams, too. We like to pay the lowest price possible and will spend much time assuring ourselves this is accomplished. We feel the same higher demands on our money for the house, food, and bills. And as this demand increases, the amount of money left for our hobby decreases. So when money is spent, we want the best quality for the best price.

There are a number of ways to bring the cost of a product down. By using a cheaper grade of material, buying raw materials in larger quantities to obtain a better discount, by cutting the profit ratio, and by eliminating the middle man. Wilson Systems will not lower the quality of the product. In fact, we have improved the strength and quality of almost every antenna in the line. The newly designed monobanders will stay up under heavy icing conditions when others are falling apart. Wilson Systems is currently purchasing at the lowest price possible from the aluminum companies, so these methods of cost reduction are eliminated. The third method mentioned is one that we have decided to consider as a part of the overall cost reduction plan, yet leaving room for research and development expense, so we may bring you the products you want and at a price you will like.

The last method mentioned is always a risky one. The dealers do not want their profits cut back just as you do not want your pay check cut. If you cut the dealers' profits back, some of them will just push the product that will tend to give them the most profit, rather than the one that will be the best performing for you. A rather drastic form of this method is the one that Wilson Systems will be choosing. You will not be able to find the Amateur products of Wilson Systems in stock at the dealers, nor will they probably recommend them. (After all, as long as they're not handling them, why should they promote or even recommend them?) No, you will only be able to enjoy the most product for the least money by dealing with Wilson Systems factory direct. We will be offering you the amateur antennas and towers at prices that are below, in most cases, what the dealers pay for the products of other companies. And to make it even easier, we have a toll-free number for you to place your order. Now isn't this what you've been looking for? The best product for the least money!

Just remember these four points:
1. Highest Quality 2. Lowest Price 3. Toll-Free Order Number

The fourth point? Remember the name . . . WILSON SYSTEMS, INC.

Yours Truly,
Jim Wilson
Wilson Systems, Inc.

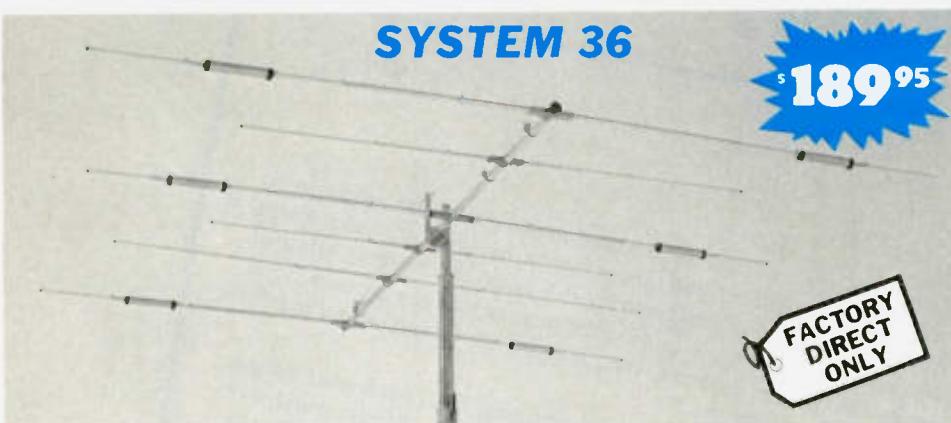
WSI **WILSON**
SYSTEMS, INC.

4286 S. Polaris Ave., Las Vegas, Nevada 89103
(702) 739-7401 — Toll-Free Order Number 800-634-6898

WILSON SYSTEMS INC. MULTI-BAND ANTENNAS

SYSTEM 36

\$189.95



FACTORY
DIRECT
ONLY

A trap loaded antenna that performs like a monobander! That's the characteristic of this six element three band beam. Through the use of wide spacing and interlacing of elements, the following is possible: three active elements on 20, three active elements on 15, and four active elements on 10 meters. No need to run separate coax feed lines for each band, as the

bandswitching is automatically made via the High-Q Wilson traps. Designed to handle the maximum legal power, the traps are capped at each end to provide a weather-proof seal against rain and dust. The special High-Q traps are the strongest available in the industry today.

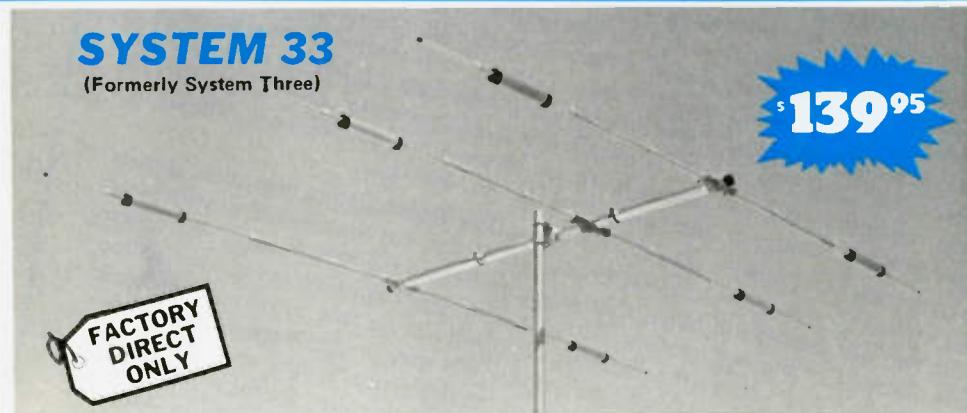
SPECIFICATIONS

Band MHz	14-21-28	Boom (O.D. x Length)	2" x 24'2½"	Wind loading @ 80 mph	215 lbs.
Maximum power input	Legal limit	No. of elements	6	Maximum wind survival	100 mph
Gain (dBd)	Up to 9 dB	Longest element	28'2½"	Feed method	Coaxial Balun (supplied)
VSWR @ resonance	1.3:1	Turning radius	18'6"	Assembled weight (approx.)	.53 lbs.
Impedance	50 Ω	Maximum mast diameter	2"	Shipping weight (approx.)	.62 lbs.
F/B ratio	20 dB or better	Surface area	8.6 sq. ft.		

SYSTEM 33

(Formerly System Three)

\$139.95



FACTORY
DIRECT
ONLY

Capable of handling the Legal Limit, the "SYSTEM 33" is the finest compact tri-bander available to the amateur.

Designed and produced by one of the world's largest antenna manufacturers, the traditional quality of workmanship and materials excels with the "SYSTEM 33".

New boom-to-element mount consists of two 1/8" thick formed aluminum plates that will provide more clamping and holding strength to prevent element misalignment.

SPECIFICATIONS

Band MHz	14-21-28	Boom (O.D. x length)	2" x 14'4"	Wind loading at 80 mph	114 lbs.
Maximum power input	Legal limit	No. elements	3	Assembled weight (approx.)	37 lbs.
Gain (dBd)	Up to 8 dB	Longest element	27'4"	Shipping weight (approx.)	42 lbs.
VSWR at resonance	1.3:1	Turning radius	15'9"	Direct 52 ohm feed—no balun required	
Impedance	50 ohms	Maximum mast diameter	2" O.D.	maximum wind survival	100 mph
F/B ratio	20 dB or better	Surface area	5.7 sq. ft.		

Superior clamping power is obtained with the use of a rugged 1/4" thick aluminum plate for boom to mast mounting.

The use of large diameter High-Q Traps in the "SYSTEM 33" makes it a high performing tri-bander and at a very economical price.

A complete step-by-step illustrated instruction manual guides you to easy assembly and the lightweight antenna makes installation of the "SYSTEM 33" quick and simple.

\$9.95

GR-1

The GR-1 is the complete ground radial kit for the WV-1A. It consists of: 150' of 7/14 stranded copper wire and heavy duty egg insulators, instructions. The GR-1 will increase the efficiency of the GR-1 by providing the correct counterpoise.

Prices and specifications subject to change without notice.

\$44.95

WV-1A

4 BAND

TRAP VERTICAL
(10 - 40 METERS)

No bandswitching necessary with this vertical. An excellent low cost DX antenna with an electrical quarter wavelength on each band and low angle radiation. Advanced design provides low SWR and exceptionally flat response across the full width of each band.

Featured is the Wilson large diameter High-Q traps which will maintain resonant points with varying temperatures and humidity.

Easily assembled, the WV-1A is supplied with a hot dipped galvanized base mount bracket to attach to vent pipe or to a mast driven in the ground.

Note:

Radials are required for peak operation.
(See GR-1 below).

SPECIFICATIONS:

- Self supporting—no guys required.
- Input Impedance: 50 Ω
- Powerhandling capability: Legal Limit
- Two High-Q Traps with large diameter coils
- Low Angle Radiation
- Omnidirectional performance
- Taper Swaged Aluminum Tubing
- Automatic Bandswitching
- Mast Bracket furnished
- SWR: 1.1:1 or less on all Bands

W S I WILSON
SYSTEMS, INC.

4286 S. Polaris Avenue
Las Vegas, Nevada 89103
(702) 739-7401
Factory Direct Toll Free 1-800-634-6898

New, Improved Wilson Towers



Hinged Base Plate - Concrete Pad, Heavy Duty Winch



Mounting the House Bracket



The Hinged Base Plate allows tower to be tilted over for access to antenna and rotor from the ground.



FEATURES:

- Maximum Height 45' (will handle 10 sq. ft. at 38') @ 50 mph
- 800 lb. winch
- Totally freestanding with proper base
- Total Weight, 189 lbs.

The TT-45A is a freestanding tower, ideal for installations where guys cannot be used. If the tower is not being supported against the house, the proper base fixture accessory must be selected.

GENERAL FEATURES

All towers use high strength heavy galvanized steel tubing that conforms to ASTM specifications for years of maintenance-free service. The large diameters provide unexcelled strength. All welding is performed with state-of-the-art equipment. Top sections are 2" O.D. for proper antenna/rotor mounting. A 10' push-up mast is included in the top section of each tower. Hinge-over base plates are standard with each tower. The high loads of today's antennas make Wilson crank-ups a logical choice.

FACTORY DIRECT
\$399.95

NEW IMPROVED FEATURE

Heavier wall tubing greatly increases the stress capabilities over the older TT-45 and MT-61.



FEATURES:

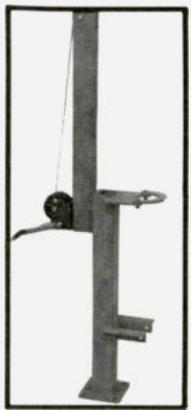
- Is freestanding with use of proper base
 - Maximum Height is 61' (will handle 10 sq. ft. at 53') @ 50 mph
 - 1200 lb. brake winch
 - 4200 lb. raising cable
 - Total Weight, 350 lbs.
- Recommended base accessory: RB-61A, FB-61A.

The MT-61A is our largest and tallest freestanding tower. By using the RB-61A rotating base fixture the MT-61A is ideally suited for the SY33 or SY-36. If you plan to mount the tower to your house, caution should be taken to make certain the eave is properly reinforced to handle the tower. If not, one of the base accessory fixtures should be used.

TILT-OVER BASES FOR TOWERS

FIXED BASE

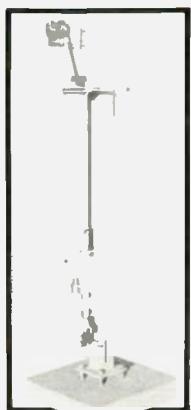
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FB-45A ... \$ 79.95
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Tilting the tower over is a one-man task with the Wilson bases.
(Shown above is the RB-61A.)

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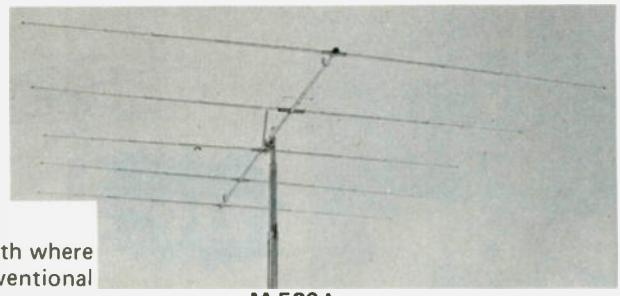
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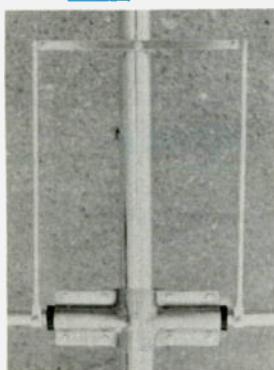
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M-520A



Wilson's Beta match offers maximum power transfer.

With the Wilson Beta-match method, it is a "set it and forget it" process. You can now assemble the antenna on the ground, and using the guidelines from the detailed instruction manual, adjust the tuning of the Beta-match so that it will remain set when raised to the top of the tower. The Wilson Beta-match offers the ability to adjust the terminating impedance that is far superior to the other matching methods including the Gamma match and other Beta-matches. As this method of matching requires a balanced line, it will be necessary to use a 1:1 balun, or RF choke, for the most efficient use of the HF Monobanders.

The Wilson Monobanders are the perfect answer to the Ham who wants to stack antennas for maximum utilization of space and gain. They offer the most economical method to have more antenna for less money with better gain and maximum strength. Order yours today and see why the serious DXers are running up that impressive score in contests and number of countries worked.

SPECIFICATIONS

Model	Band Mtrs	Gain dBD	F/B Ratio	Bandwidth @ Resonance 2:1 VSWR Limits	VSWR @ Resonance	Impedance	Matching	Elements	Longest Element	Boom O.D.	Boom Length	Turning Radius	Surface Area (Sq.Ft.)	Windload @ 80 mph (Lbs.)	Maximum Mast	Assembled Weight (Lbs.)
M520A	20	11.5	25 dB	500 KHz	1.1:1	50 Ω	Beta	5	36'6"	2"	34'2½"	25'1"	8.9	227	2"	68
M420A	20	10.0	25 dB	500 KHz	1.1:1	50 Ω	Beta	4	36'6"	2"	26'0"	22'6"	7.6	189	2"	50
M515A	15	12.0	25 dB	400 KHz	1.1:1	50 Ω	Beta	5	25'3"	2"	26'0"	17'6"	4.2	107	2"	41
M415A	15	10.0	25 dB	400 KHz	1.1:1	50 Ω	Beta	4	24'2½"	2"	17'0"	14'11"	2.1	54	2"	25
M510A	10	12.0	25 dB	1.5 MHz	1.1:1	50 Ω	Beta	5	18'6"	2"	26'0"	16'0"	2.8	72	2"	36
M410A	10	10.0	25 dB	1.5 MHz	1.1:1	50 Ω	Beta	4	18'3"	2"	12'11"	11'3"	1.4	36	2"	20

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	M-520A	5 Elements on 20 Mtrs.	TRUCK	199.95		RB-61A	Rotating Base for MT-61A w/tilt over feature	TRUCK	179.95
	M-420A	4 Elements on 20 Mtrs.	UPS	139.95		FB-61A	Fixed Base for MT-61A w/tilt over feature	TRUCK	109.95
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	M-415A	4 Elements on 15 Mtrs.	UPS	79.95					
	M-510A	5 Elements on 10 Mtrs.	UPS	84.95					
	M-410A	4 Elements on 10 Mtrs.	UPS	64.95					
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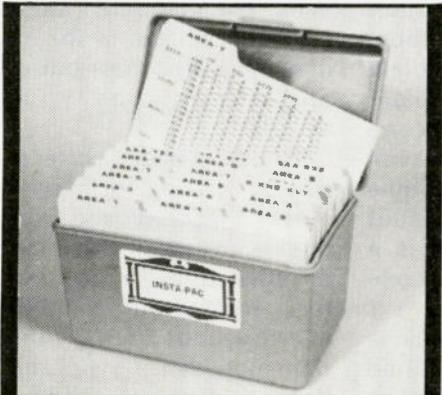
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No More TRS-80 Cassette Woes

— E-Z Loader does the trick

Paul L. Goetz WA9PUL
2228 Madison Place
Evanston IL 60202

David F. Miller K9POX
7462 Lawler Ave.
Niles IL 60648

You can eliminate most of your Level II TRS-80 cassette-loading problems once and for all! It's true, the circuit described in this article will go a long way toward eliminating your CLOAD problems. If all of the data is there on the cassette, this gadget will find it and load it into your TRS-80 Level II machine... it's the greatest thing since the Level I manual!

If you've worked with Level II for any time at all, you know that it is somewhat sensitive to playback volume on a CLOAD command (what an understatement!). Playing back a good recording through this circuit—let's call it the E-Z Loader—will allow you to set the playback volume on the Radio Shack CTR-41 recorder anywhere from 2 to 10! A not-so-good recording will have some-

what less range, but it will still be quite broad, and you will be able to load it. Even the head azimuth (vertical alignment of the record vs the playback head if not the same machine) becomes fairly noncritical, and you can even vary the playback volume up and down while loading with no adverse effect.

Sound too good to be true? It seems that in this day and age we're besieged with all manner of claims and counterclaims, but E-Z Loader really works... try it and see for yourself!

The Problem

When you type CSAVE and ENTER, the TRS-80 puts out a positive- and negative-going square wave at the cassette port. However, when the audio recorder "processes" this square wave and lays it down on tape, it isn't exactly square any longer. Take a peek at Fig. 1 and see what we mean!

The output of the cassette recorder as depicted in Fig. 1(b) is more like a mostly negative-going spike, broadened out, with a small (in ratio) positive dimple on either side. This is what a good recording looks like; a bad one is downhill from that! Somehow, the TRS-80 will ac-

cept that waveform and load the data it contains into RAM, if the recorder/playback head azimuth is quite close and the playback level is just right!

This is where the circuit of Fig. 2 comes into action. The root of the problem lies in the TRS-80's resident software. After the computer recognizes a sync pulse, it waits a short time and then looks at the input once again. If there is a pulse present, the computer interprets it as a 1 and goes through the cycle all over again. If there is no pulse present, the computer assumes this to be a 0 and repeats the cycle.

This sequence occurs until all of the data has been loaded from the tape, at which time the READY prompt reappears. The problem in all of this, however, is in the amount of time that the TRS-80 waits after recognizing a pulse, before it looks for the next pulse, and not in the input level, as many believe!

Currently, a Level II TRS-80 waits only about 250 microseconds before testing for a pulse, and therein lies the problem. A good tape will produce a pulse that is much narrower than 250 microseconds—see Fig. 1(b)—but a bad tape, or one in which the head azimuth of the

record machine differs from the playback machine, can easily produce pulse widths in excess of that magical 250 microsecond figure! When that happens, the computer interprets the tail end of the initial pulse as a new pulse, and there goes your CLOAD!

The E-Z Loader corrects this by waiting for a sync pulse, triggering and producing an output pulse 150 microseconds wide and then locking-out for 750 microseconds (2/3 of the time between a sync and data pulse), triggering again if a data pulse (logic 1) is present, and outputting it as a 150-microsecond-wide pulse. Thus, no matter how wide your input pulses are, they can produce only 150-microsecond-wide output pulses.

In a nutshell, the E-Z Loader circuitry takes the rather poorly defined spike of Fig. 1(b) and regenerates it back into a nice square, narrow pulse again. Your computer will really love that narrow pulse and will

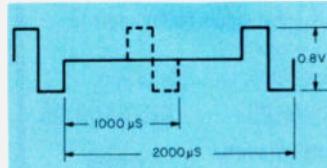


Fig. 1(a). Output directly from the cassette port of the computer. Sync pulses are solid lines; data pulses are dotted lines.

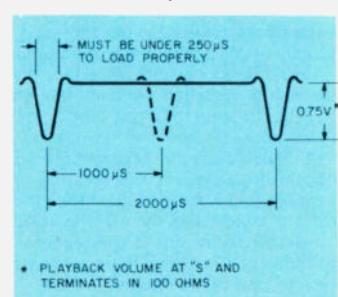


Fig. 1(b). The output from a good cassette tape after being recorded by the CTR-41 recorder. Sync pulses, solid; data pulses, dotted.

continue to reward you for your effort.

How the Circuit Works

Following the schematic diagram (Fig. 2) from the cassette player output (which is the input to the E-Z Loader), you'll notice that we transformer-couple in order to make the bridge rectifier behave properly and to eliminate the ground loop condition that exists in the Radio Shack CTR-41 recorder (in case you haven't eliminated it already by other means).

The input signal to the transformer first passes through a high-pass filter (C_5 and R_{10}) and is stepped up by a factor of at least 10. It is then applied to the bridge rectifier whose output consists of only positive-going pulses. These positive pulses are sent to the base of transistor Q_1 , but with a slight twist.

Locate R_7 and C_4 on the schematic and notice that Q_1 's base current must pass through C_4 . As pulses are rectified by the bridge, C_4 charges to a voltage equal to the output of the bridge (less 0.6 V) and at the same time supplies a current spike to Q_1 's base, triggering the transistor. By the next time a pulse arrives, C_4 has discharged slightly through R_7 , and again supplies a triggering current spike to Q_1 's base.

The net effect of all of this is that Q_1 , C_4 , and R_7 form a passive level detector and trigger the rest of the circuitry only on pulse peaks, regardless of the level arriving from the recorder. The amplified and peak-detected pulses are fed to the first inverting input of the 74LS123 IC, which is a dual low-power Schottky retriggerable monostable multivibrator chip (say that three times!).

The two multivibrators in this package are connected in such a way that

the output is taken from the Q output of section one, which also feeds the non-inverting input of section two. The \bar{Q} output of section two is fed back to the non-inverting input of section one, which has the effect of disabling the output of section one for a 750-microsecond duration.

The total effect of all of this can be observed on an oscilloscope at pin 13 (or pin 10) of the IC: a positive-going square pulse, 150-microseconds wide, for every sync and data pulse leaving the tape, regardless of the width of the pulse on the tape. No hum, no noise, no miscellaneous crud to fool the TRS-80 into loading erroneous data... just nice, clean pulses. The buffer amplifier is an emitter-follower with a voltage gain somewhat less than unity, but it does a nice job of isolating and matching the 100-Ohm input impedance of the TRS-80 computer.

TRS-80 performance won't be affected by positive, negative or square waves (both transitions), so we chose positive pulses to keep the circuit parts count down. The pulses output to the computer are between 1 and 2 volts peak-to-peak (viewed on a scope with a 100-Ohm termination), regardless of where the CTR-41 recorder volume control is set, as long as there is at least enough volume to detect the pulses and trigger the E-Z Loader's circuitry. If the playback volume is too low, there simply will be no output (or possibly, sporadic output).

The point of minimum output for each cassette is easy to determine, however, and as long as you stay above that point, you'll get a good load. On a decent tape, this minimum can be 2 or 3 on the CTR-41's graduated volume control knob, and there

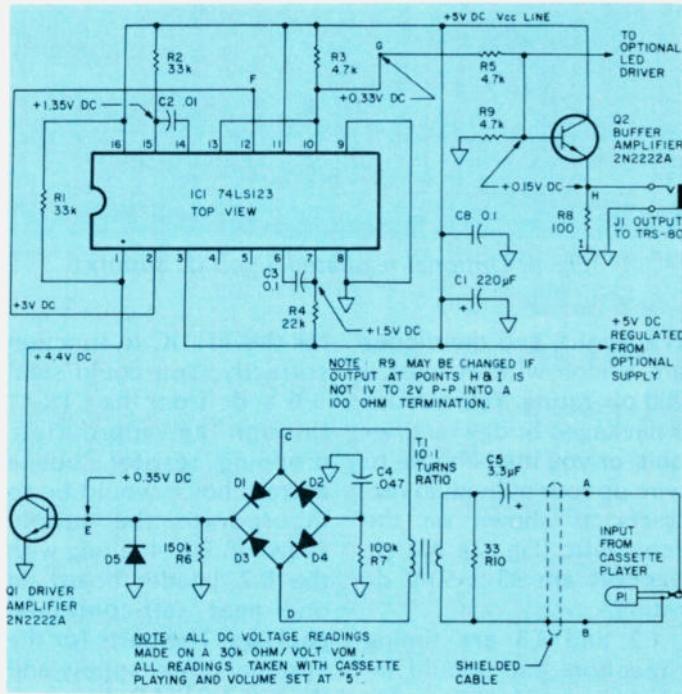


Fig. 2. TRS-80 E-Z Loader schematic diagram.

usually is no maximum level.

Precautions

One potential pitfall to properly loading with the E-Z Loader in-line (other than tape dropouts and/or bad data on the tape) would be insufficient level into the E-Z Loader, thus the optional LED driver shown in Fig. 3. The LED visually shows the operator that point at which the E-Z Loader is not receiving enough level from the recorder on playback. As long as the LED is lit and slightly flickering, the recorder output level control (volume) is set above this threshold. It also adds a nice touch in impressing visiting dignitaries.

It's an option because it has no effect on the rest of the circuitry and is for operator convenience only. To prevent occasional drops in level on the tape from causing a bad CLOAD, keep the recorder's output at least two divisions above the point at which the LED extinguishes.

Requirements

OK, we've talked you into building the E-Z Loader,

now you probably think that there will be some sort of catch such as specially selected parts, critical circuit layout, and/or fancy test equipment needed for alignment. No! All of the parts are available from Radio Shack with the possible exception of the input transformer, which is actually an output transformer with a turns ratio of approximately 10:1. You can probably scrounge one from that defunct \$9.95 portable transistor radio you've been putting off fixing. That's called recycling and it's the "in" thing to do.

The transformer is hooked up backwards so that the low impedance (speaker) winding is accepting the audio out of the cassette recorder. Resistors are all 1/4 or 1/2

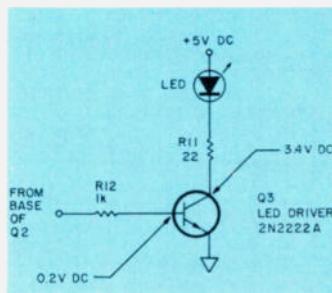


Fig. 3. Optional LED driver circuit.

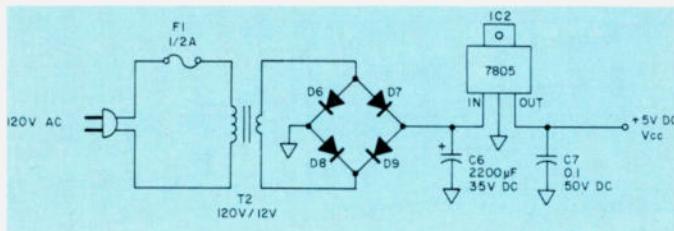


Fig. 4. Optional regulated 5-volt dc supply.

Watt units, and the diodes are silicon with at least a 200 piv rating. You can use a packaged bridge rectifier unit, or you may choose to wire up four individual rectifiers as shown on the schematic (Fig. 2). All capacitors are 35-to-50-V dc rating.

C2 and C3 are timing capacitors and should be polyester or tantalum for stability and accuracy of value. The 2N2222A transistors are medium-gain devices and also are not critical. Whether you use perfboard layout (as shown in the photo) or a printed circuit board, it would be advisable to consider using a socket for the 74LS123 IC (I just hate unsoldering 16 pins all at once).

As you may have noticed, the E-Z Loader requires +5 V dc (at about 25 mA), and it should be quite close to that figure

for the TTL IC to function correctly. You could steal +6 V dc from the CTR-41 through an appropriate dropping resistor, but a better choice would be to incorporate the supply shown in Fig. 4, along with the E-Z Loader board, in one neat self-contained package. The parts for the optional power supply and the optional LED driver are also available from Radio Shack. Now you don't have any excuse for putting off building the gadget!

Putting E-Z Loader to Use

There is no alignment or parts selection required. Voltage readings and oscilloscope waveforms at various points are shown to assist in verifying that the unit is functioning properly. Simply hook up the E-Z Loader between your CTR-41 recorder's "ear" output and the TRS-80's audio input and you're all set. Incorporating a 1/8-inch phone jack on its output and a 1/8-inch phone plug on its input will permit you to remove the E-Z Loader at any time and normalize your setup. But we honestly don't think that you'll be inclined to do that very often, not once you've seen how easy it is to CLOAD with the E-Z Loader in-line (except, perhaps, to demonstrate the difference to skeptical friends).

One other use for the E-Z Loader would be to clean up poor quality tapes. Using a second recorder, play the bad tape through the

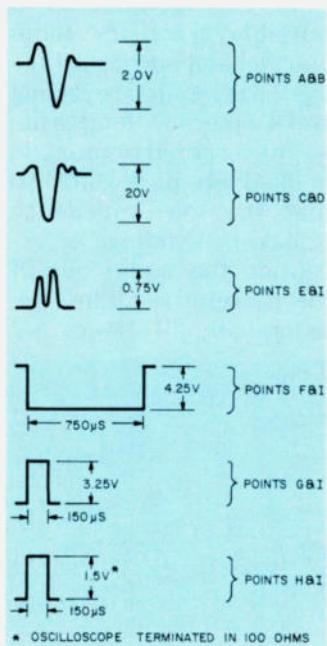
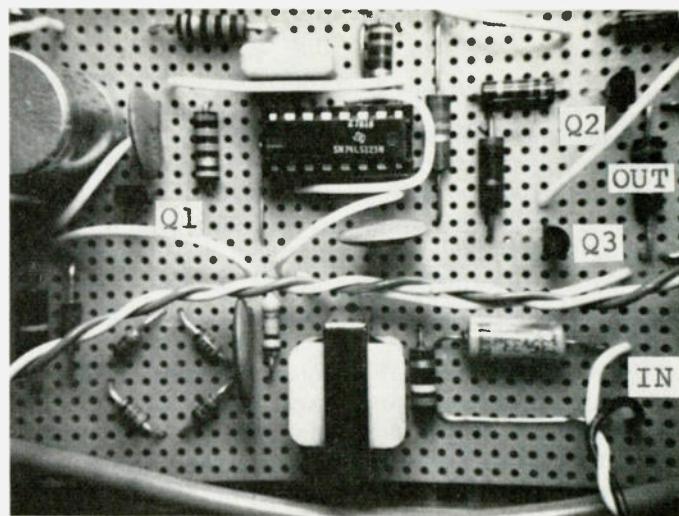


Fig. 5. Peak-to-peak oscilloscope measurement within the E-Z Loader's circuitry.



The perfboard layout used in one of the units that has been constructed.

E-Z Loader and plug the E-Z Loader's output into the second recorder's aux input. The recording on the second machine will then be first quality, and should load on anyone's unmodified computer with ease.

Exceptionally high hum

level on the tape being played back and wow (frequency instability) in the record or playback machines are other potential sources of CLOADing errors. The E-Z Loader helps in all cases, but obviously there are limits beyond

Parts List

- C1—220- μ F electrolytic, R.S. #272-1029
- C2—.01- μ F printed circuit capacitor, R.S. #272-1065
- C3—1- μ F tantalum capacitor, R.S. #272-1401
- C4—.047- μ F printed circuit capacitor, R.S. #272-1068
- C5—3.3- μ F tantalum capacitor, R.S. #272-1408
- C6—2200- μ F electrolytic, R.S. #272-1020
- C7, C8—.1- μ F ceramic disc, R.S. #272-135
- D1 to D9—200-volt silicon rectifiers, R.S. #276-1102
- F1—1/2-Amp, 1/4 x 1-1/4 inch fuse, R.S. #270-1271 mounted in R.S. #270-364 holder
- IC1—74LS123 integrated circuit, R.S. #276-1926
- IC2—7805 integrated circuit, R.S. #276-1770
- J1—1/8-inch phone jack, R.S. #274-251
- P1—1/8-inch phone plug, R.S. #274-286
- Q1 to Q3—2N2222A transistor, R.S. #276-1617
- R1, R2—33k, 1/2-W, 10 percent resistor, R.S. #271-040
- R3, R5, R9—4.7k, 1/2-W, 10 percent resistor, R.S. #271-030
- R4—22k, 1/2-W, 10 percent resistor, R.S. #271-038
- R6—150k, 1/2-W, 10 percent resistor, R.S. #271-047
- R7—100k, 1/2-W, 10 percent resistor, R.S. #271-045
- R8—100-Ohm, 1/2-W, 10 percent resistor, R.S. #271-012
- R10—33-Ohm, 1/2-W, 10% resistor, R.S. #271-007
- R11—22-Ohm, 1/2-W, 10% resistor, R.S. #271-005
- R12—1k, 1/2-W, 10% resistor, R.S. #271-023
- T1—audio output transformer, 10:1 ratio or greater, such as Callectro #D1-729 (200-Ohm c-t to 8 Ohm)
- T2—120 V ac to 12 V ac @ 300-mA transformer, R.S. #273-1385

Miscellaneous

- Socket for IC1—16-pin DIP, R.S. #276-1998
- LED—light emitting diode, R.S. #276-1622
- Solder, hook-up wire, shielded cable, perfboard and cabinet
- Note: C2 and C3 are timing capacitors and as such must be of good quality such as those listed above. Do not use disc types for this purpose.
- A circuit board for the E-Z Loader and power supply is available from the authors. Send \$6 and an SASE to Paul Goetz, 2228 Madison Pl., Evanston IL 60202.

which very little can be done.

Keeping your recorder's head, pinch roller, capstan and tape guides clean will go a long way toward helping the E-Z Loader do its job. Using fairly decent cassettes without wrinkles on the tape, fixing sticking reels and poorly placed pressure pads, and watching out for nonuniform oxide coating are also your responsibilities toward

good loading. As stated in the beginning of the article, as long as the data is on the tape, the E-Z Loader will go a long way toward digging it out and loading it into your TRS-80, but you will have to give it a fair shake at least!

It has been our intent to make public a circuit that is genuinely needed within the TRS-80 community. We intend no infringement on the rights of others, and

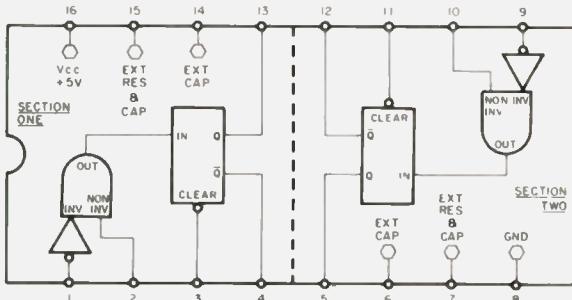
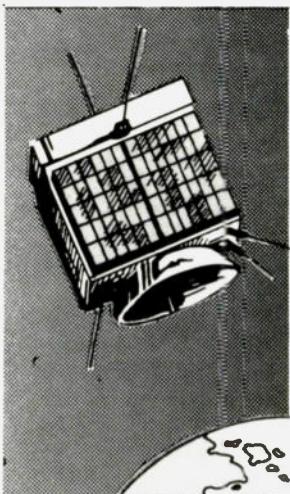


Fig. 6. Internal logic for the 74LS123 chip.

reproduction of the circuit is granted on a not-for-profit basis only. I will answer

all inquiries if the writers would be kind enough to enclose an SASE. ■



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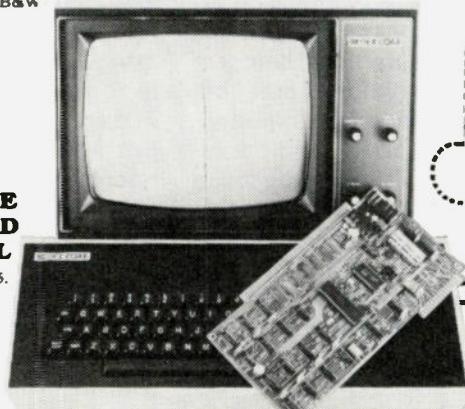
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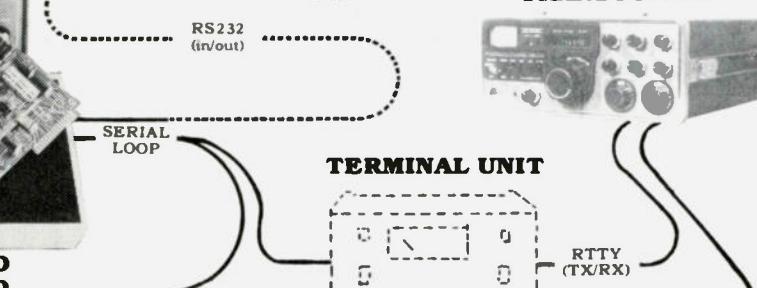


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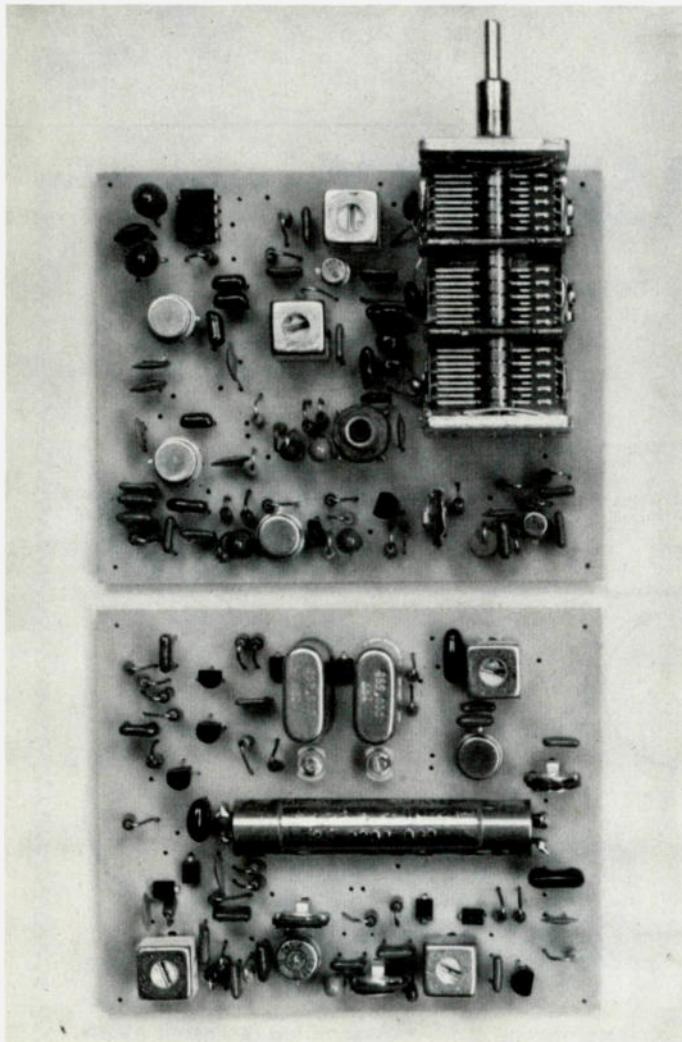
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Over the years, a ham's junk box can sport a pretty wild assortment of goodies, and mine is no exception. Many of the items are treasured for their sentimental value, some because they are of the often-needed variety, and the remainder simply because they cost a bunch. In this last category, I had stored away several Collins mechanical filters which by now had been with me so long that I couldn't remember how I had acquired them in the first place. On several occasions I had tried to sell or swap them, but no takers had appeared, so I decided I might just as well put them to use.

A Miniature Transceiver

If you are familiar with my Minicom series of receivers, you know I have a passion for miniaturization

and it would seem logical that the filters be used in something small. The final choice was an SSB generator and receiver combination which would ultimately form the major portion of a transceiver. The entire project occupies 2 PC boards of identical size. Both boards are 3.7" x 3" and can be stacked if desired. The receiver portion is a simplified version of the Minicom MK V and covers 3.5 to 4.0 MHz. The transmitting exciter section output is also 3.5 to 4.0 MHz. By adding a suitable amplifier, the rig could be used as is on 75 meters. For use on other bands, additional mixers and a crystal oscillator would be used to heterodyne up to the desired frequencies.

Circuit Description

The Collins filters from my junk box are all housed

in the Y-style case which is cylindrical and for which the PC layout is designed. Of the 2 filters from the collection which were suitable for this application, one had a bandwidth of 2 kHz and the other 3.1 kHz. Both worked well in this circuit.

A complement of 6 integrated circuits and 8 transistors provide all the needed functions. One i-f stage, the vfo, and the bfo are common to both receive and transmit modes. The input to the mechanical filter and the output from the common i-f stage are transferred from receive to transmit by means of diode switches. The upper- and lower-sideband crystals in the bfo are also switched by diodes.

All the frills such as VOX, noise blower, CW filter, S-meter, and other goodies were left out in the interest of miniaturization. When assembled sandwich fashion, the whole rig fits in the palm of your hand.

Speech Amplifier and Balanced Mixer

A CA3020 integrated circuit performs as both speech amplifier and balanced mixer, thus contributing substantially to our miniaturization efforts. This device, though designed for class B audio amplifier service, has a bandwidth of 8 MHz and lends itself to rf applications such as this. The chip houses an emitter-follower input stage, a differential amplifier, 2 emitter-follower drivers, and 2 output transistors. The output transistors have uncommitted collectors and emitters, which makes the device suitable for use in our circuit.

The mike feeds into pin 10 which is the base of the input emitter-follower. Input impedance is over 50k Ohms. R2 is a preset trimmer used to adjust the audio gain for the mike be-

ing used and, once set, can be left alone. The 455-kHz carrier is introduced across the emitters of the output transistors and a 50-Ohm

preset pot is used to balance the signal. In some cases, cancellation of the carrier from the output may not be complete and

capacitive balancing may be necessary. The PC board has provision for a small capacitor on either side of the balance pot to ground.

A modified transistor i-f transformer is used to couple the collectors of the output transistors to the mechanical filter. The winding data for this and other transformers will be covered later.

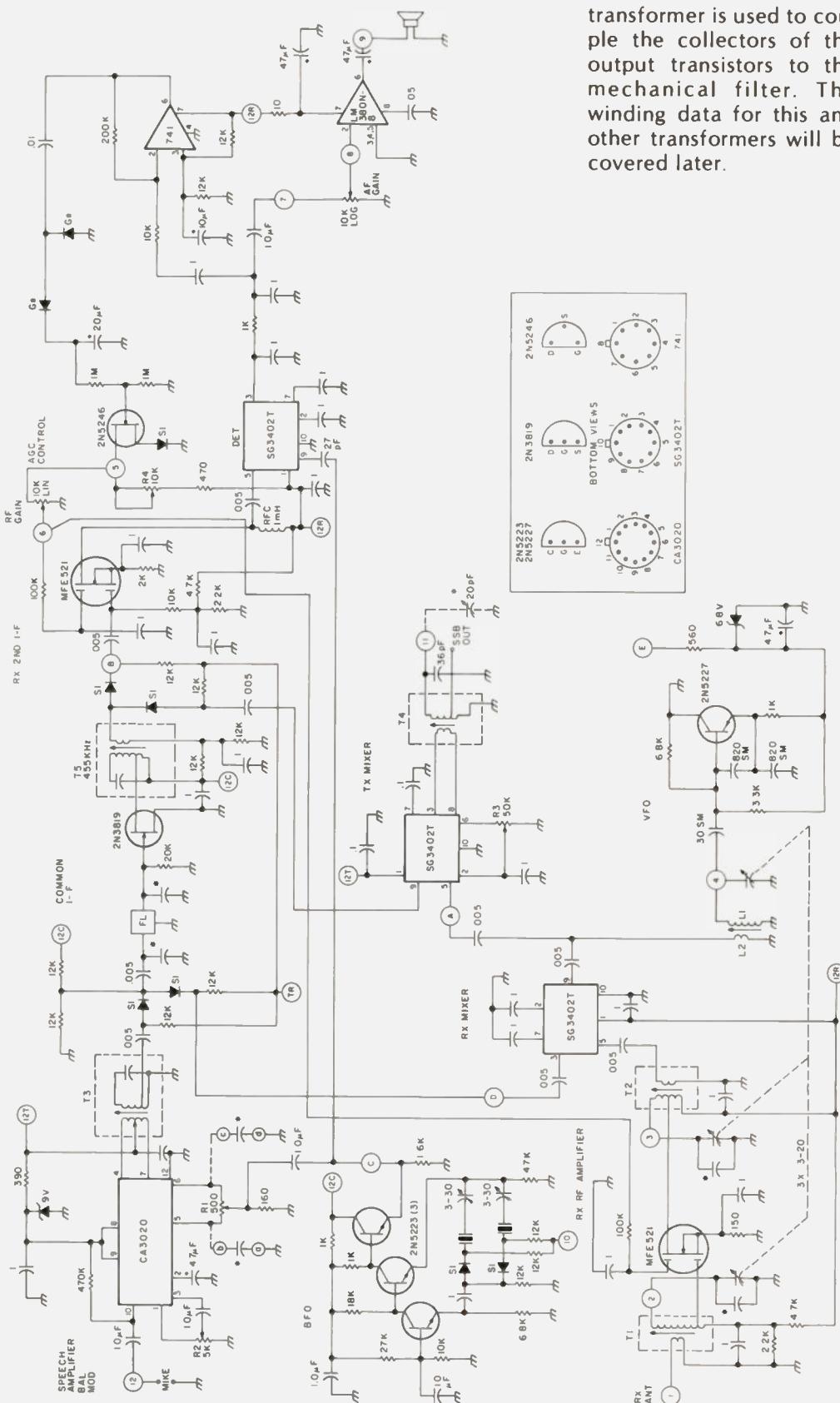


Fig. 1. Complete schematic for the transceiver. Circled letters and numbers refer to pads on the PC boards. *See text. Si = silicon diode. Ge = germanium diode.

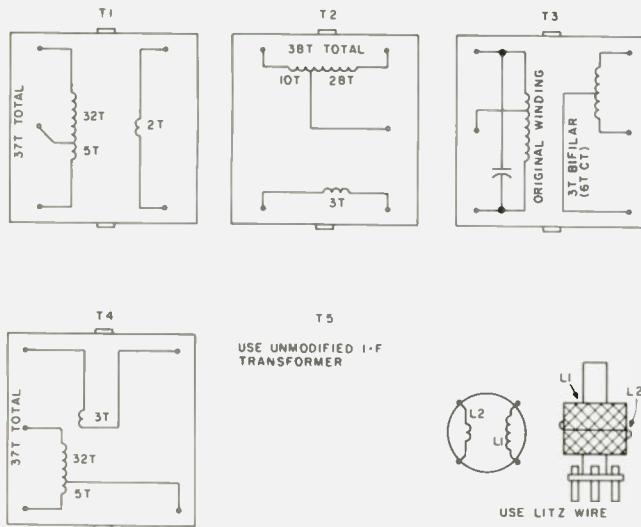


Fig. 2. Coil winding table. All bottom views.

I-f Amplifier

The SPDT diode switch at the input to the Collins filter connects it either to the transmitter mixer or to the receiver mixer output. Following the filter is a JFET amplifier that stays in the circuit for both transmitting and receiving. Then comes a second SPDT diode switch that routes the output to a second i-f stage for receiving or to a mixer for transmitting. The second i-f stage uses a dual gate FET with agc applied to gate 2.

Bfo, Vfo, and Transmitting Mixer

The bfo is crystal controlled and operates continuously since it is common to both receive and transmit modes. An SPDT diode switch selects the desired crystal for either upper- or lower-sideband operation.

The vfo tunes from 3.045 to 3.545 MHz. It is also a common circuit and operates continuously. It is actually a part of the receiver section and is the same circuit that has been used in the Minicom. The common vfo and bfo ensure that both transmitter and receiver will be on the same frequency. Since very little energy is required by the mixers, no buffers are used after the vfo. Output

is taken from a 1-turn link wound over the tank coil.

The 455-kHz signal and the output from the vfo are mixed in an SG3402T IC balanced mixer. A 3.5-to-4.0-MHz output is produced and the vfo signal is nulled by means of the 50k preset trimmer, R3. The output transformer, T4, is fabricated from a standard transistor i-f transformer. A PC pad is provided for connecting a small variable capacitor of 20 to 30 pF across the coil for front-panel peaking at any frequency. A tiny solid dielectric type from a transistor FM radio is ideal for this purpose.

Receiver

All the circuits used in the receiver section are the same as the Minicom MK V. For those not familiar with the DMOS transistors, such as the Motorola MFE521 used in the rf and i-f stages, they require a fixed bias of around 4 volts on gate 1. This is provided by the 2.2k and 4.7k resistors which form a divider across the 12-volt supply. Gate 2 is controlled by agc voltage but, unlike the regular MOSFETs, need not go negative to attenuate the signal. The DMOS transistor works with all positive bias which simplifies the agc. R4 is ad-

justed so that the no-signal level on the agc line is 6 volts. The transistors will be working at close to maximum gain at this value of bias and a strong signal will drive this voltage down to 3.5 to 4 volts. A manual rf gain control is provided and would normally be mounted on the front panel.

The mixer is an SG3402T IC which provides substantial gain and, as mentioned before, places a very light load on the vfo. Output from the mixer is routed to the filter via the diode switch.

A third SG3402T is used as a product detector and feeds the LM380N-8 audio IC directly. Some of the audio output from the detector is fed to a 741 op amp where it is amplified and rectified to provide a positive dc voltage which charges the 20-uF tantalum capacitor across its output. Part of this charge is bled off by means of the two 1-meg resistors and applied to the gate of the 2N5246 agc control transistor. As the dc level increases due to strong signals, the transistor conducts more heavily and causes the voltage at the drain to drop. Since the drain is connected directly to the agc line, receiver gain is decreased and the purpose accomplished.

Filling in the Details

There are some numbered pads on the boards and some with letter designations. These notations also appear on the schematic and should aid in keeping the external wiring straight. The pads marked capital A, B, C, D, and E on both boards have to be connected together in pairs, that is, A to A, B to B, etc. Mating pads are directly above one another with the boards properly oriented and component side up. This allows stacking if desired. Pad A is vfo

output. B is output from the common i-f, and C is bfo output. Pad D is input to the filter, while E connects 12 volts to the vfo from the other board's constant 12-volt supply connected to pads marked 12C.

Pads to accommodate a balancing capacitor if it should be needed as mentioned earlier are marked with small letters. One pair is a and b, the other c and d.

Pads marked 12R are for 12 volts applied during receive mode, and 12T designates a connection for 12 volts applied during transmit. Only one pad is needed; others are spares. There is a separate pad (12C) for the bfo, however, and it should be connected to a constant source of 12 volts along with one of the other 12C pads.

The pad marked TR is used to control the diode switches in the i-f strip. A 12-volt level is used for transmit and a ground level for receive. A TR relay would normally control this line as well as the 12R and 12T connections. The bfo crystals are similarly controlled by application of either 12 volts or ground to pad 10.

Of the other numbered pads, none needs individual explanation since the schematic clearly indicates their locations and the function becomes obvious.

Coils and Transformers

All the transformers are fabricated from regular 10-mm 455-kHz transistor i-f transformers. T1, T2, and T4 require the bare parts only, which means removing the tuning capacitor and all wire from the bobbin. Carefully salvage all the wire as it will be used to wind the new transformers.

The large winding goes on first and the link is then wound over the top. T1, T2, and T4 should be wound as per the table, making sure

the pinout is correct. Views are looking at the pin end or bottom of the assembly.

For T3, carefully break off the secondary leads where they enter the bobbin. This is the side with 2 pins. Unsolder the remaining wire from the pins and clean off excess solder. Since the new winding is center-tapped, you'll have to steal a pin from a spare assembly and push it into the existing hole in the base. The new link is bifilar wound over the top of the existing winding and connected as per the table. Identify the winding in some way so you won't have trouble orienting the transformer when mounting it on the board. The new winding goes to the CA3020 output.

The vfo tank coil, L1, is pie-wound on a slug-tuned PC coil form. I used a Gowanda series 7 Velvetork form which is .209" in diameter by .625" long. A carbonyl E (red) core was used for the slug. Impregnate the winding with hot coil wax and put a single turn link over the top of the pie for L2.

Construction

I doubt that there will be any mad rush to duplicate this rig exactly as it stands, but parts of it may be of interest to some readers. I even debated the need to supply PC layouts, but decided they might prove helpful in some way. Just in case you do wish to copy the layout as much as possible, the following information may be useful.

All resistors, diodes, and rf chokes were mounted hairpin fashion to conserve space. Miniature low-voltage ceramic capacitors were used for coupling and bypass applications. Polarized capacitors are dipped tantalum and resistors are $\frac{1}{4}$ Watt.

Before mounting the SG3402T used as the receiver mixer, cut off pin 6.

Likewise cut off pin 4 of the product detector. The same goes for pin 11 on the CA3020.

The 2 trimmer capacitors in series with the bfo crystals are subminiature (5-mm) units in case you were wondering how they squeezed into the space allotted to them. Incidentally, the exact operating frequency for the crystals will depend on the particular filter used, but this circuit will pull quite a bit and allow appreciable leeway in crystal accuracy.

Note that there are 2 silver mica capacitors across the ends of the mechanical filter whose purpose it is to tune the two transducer coils. Values will depend on the type of filter used. For the F455 Q2 that I used, I had to install a 91-pF capacitor at one end and a 110-pF capacitor at the other end.

The 3-gang tuning capacitor is the same one used in the Minicom receivers and has a range of approximately 3 to 20 pF per section. An additional padding capacitor is required across each of the first 2 gangs. Use 20- to 22-pF silver micas and solder directly to the frame before installation. When mount-

ing this item, use #4-40 screws $\frac{1}{4}$ " long. Put a toothed washer under the head to allow good contact with the copper. Place 2 flat washers over each screw on the component side before mounting the capacitor. This will leave enough space to clear the rivets that hold the compression trimmers for the first 2 gangs.

I used MPN3401 (Motorola) diodes to do the

switching since they were in my junk box and are made especially for this use. Regular silicon diodes should work okay, so don't panic. No other remarks come to mind regarding components, so I'll get on with the checkout.

Tune-Up and Checkout

A DPDT toggle switch can be used for TR switching by using one section to apply 0 or 12 volts to the

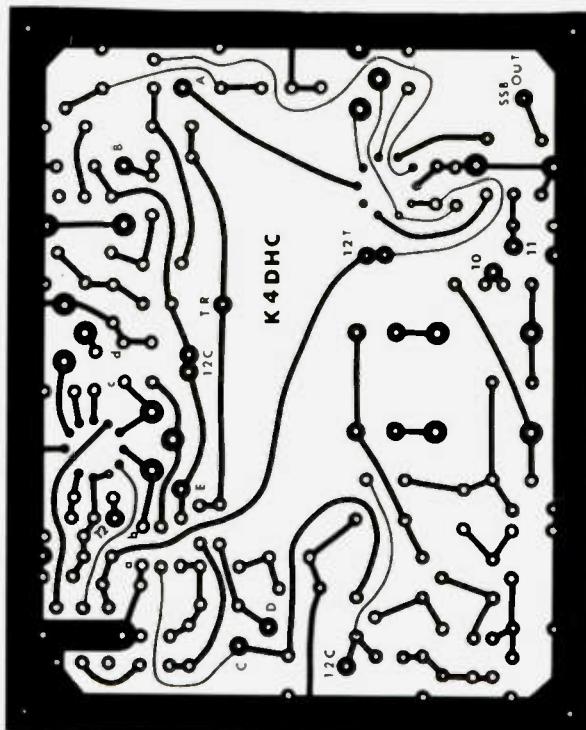


Fig. 3. PC board for transmitter.

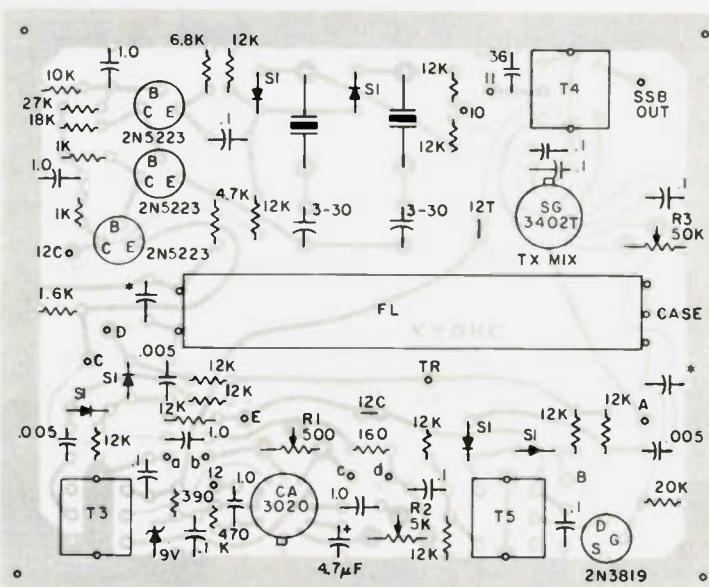


Fig. 4. Component layout for transmitter board.

TR line and the other half to transfer 12 volts between the receive and transmit circuits. An SPDT toggle can be used to switch the bfo crystals. A 10k audio gain control with log taper and a 10k rf gain control with linear taper are the only other items needed at this time. With the boards lying on the bench, make the jumpers

between boards long enough so that each board can be handled freely while making adjustments. The copper border around each PC is ground and should be made common with all other grounds and power-supply return.

Connect an 8-Ohm speaker and 12-volt power supply to the rig. Set all trimmers to midpoint. Run

the screws on the 2 compression trimmers on the 3-gang capacitor up snug but not tight. Turn the variable to full mesh. Connect a dc scope or high-impedance voltmeter between the top of the rf gain control pot and ground.

Switch to receive and apply power. Current drain should be between 55 and 65 mils. Turn the rf gain control all the way down and adjust R4 for a reading of 6 volts. Disconnect the meter and turn the rf gain to maximum. Turn up the audio gain and feed in some signal at 3.5 MHz. Adjust the slug in the vfo tank coil until the signal is picked up. Adjust the cores in T1 and T2 for maximum output. Run the variable capacitor up to the high end and feed in signal at 4.0 MHz. Peak the response by means of the 2 compression trimmers on gangs 1 and 2. Repeat these 2 procedures as many times as needed to achieve good tracking. T5 can be peaked on the noise present. If the band is active, you can connect an antenna and listen on the air to verify proper operation of the receiver. Also make

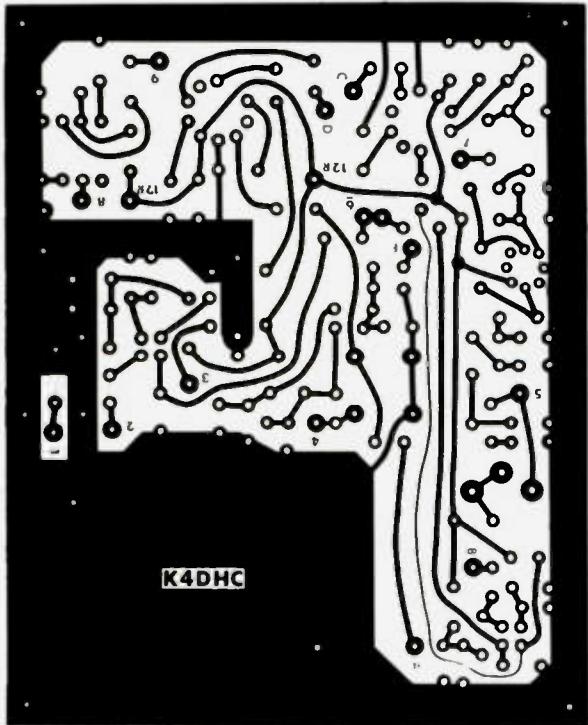


Fig. 5. PC board for receiver.

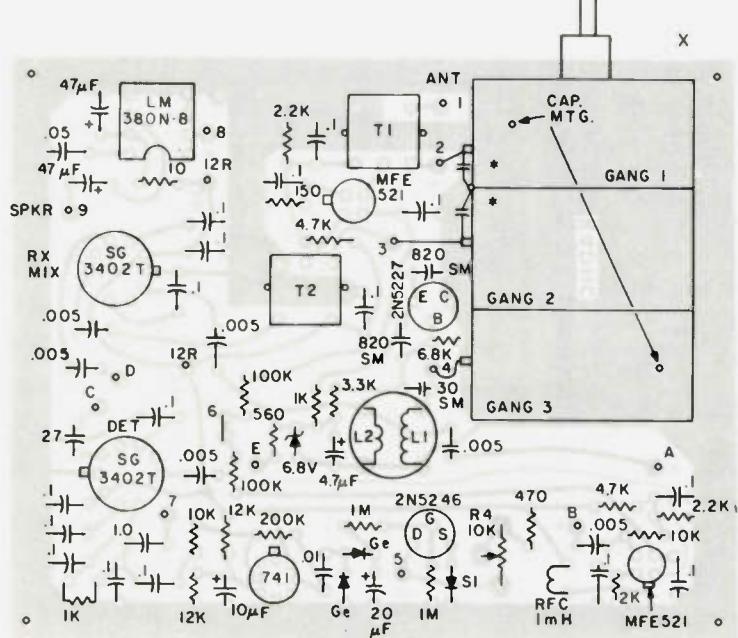


Fig. 6. Component layout for receiver board.

sure both bfo crystals are working. This concludes receiver checkout.

An audio generator should be connected to the mike input during tune-up of the transmitter. Turn R2 to minimum and switch to transmit mode. Current drain should be 40 to 50 mils. Connect a scope to the output side of T3. A piece of discarded pigtail can be temporarily soldered to the pad if necessary to make the scope connections. With no audio input, adjust R1 to null out the carrier. If there is still some trace of rf, tack a mica trimmer across pads a and b and then c and d to see where improvement can be made. Measure the value of the trimmer at the best setting and substitute a silver mica fixed capacitor.

Once the carrier has been nulled, turn up R2 and feed in some audio signal at around 1500 Hz. Adjust the level to a point just below where distortion occurs as observed on the scope. Transfer the scope to the output side of T5. Determine the values for the capacitors that tune the filter and install them at this time. Turn the tuning capacitor up to the high end of the band just below 4.0 MHz and transfer the scope to the transmitter output at T4. Shut off the signal generator and null the vfo signal by adjusting R3. Turn on the generator and adjust the core in T4 for maximum output. Later on, a tuning capacitor can be mounted on the front panel to allow peaking of the output as frequency is changed. That concludes tune-up of the transmitter.

I've listened to the signal from this little rig on some of my Minicomms and can report that it sounds excellent.

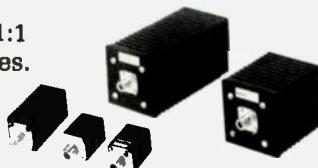
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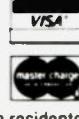
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In Quest of Perfect Break-In

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PROLOGUE: About twenty years ago—no, more—it was the year the Russians blew our minds and our national ego by putting up their *Sputnik 1*—I found an article in an encyclopedia in the library at my high school.¹ The essence of the piece was that just anybody could have his own private radio station and communicate with other anybodies who had *their* own private radio stations. It gave detailed instructions for building a one-tube receiver. It hinted at a simple transmitter and even let on that one should get a license before transmitting.

I sent off for a hundred-pound grab-box of electronic goodies. I used only one thing: a filament transformer to heat the single filament of that single tube in my first home-brew receiver. I wound one of the coils a few turns too much (using a cardboard core from a roll of paper towels), and, behold! The local broadcast stations could be tuned in. A veritable miracle on a masonite frame.

Next, from another article, I built a simple transmitter. No, I didn't get the license, and I didn't even learn the code. But I didn't break the law, either, al-

though I intended to. This little widow-maker, I later learned, was high-voltage keyed. When I fired it up, it knocked me across the room as soon as I touched the key. I took it apart and let my yen for a personal radio communications station cool for fifteen or twenty years.

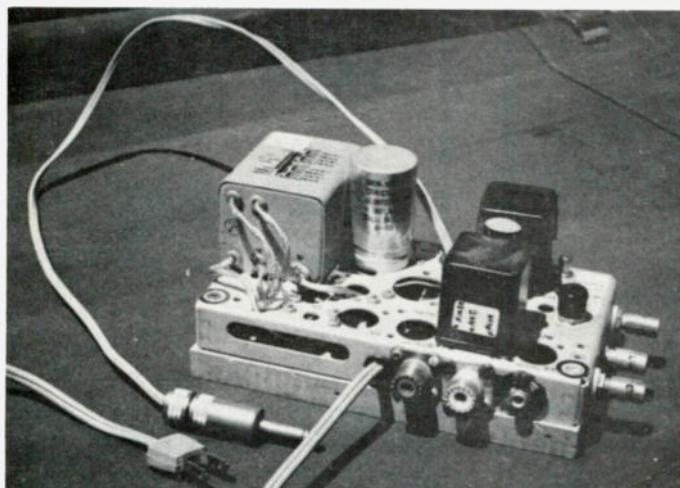
After I got out of the Army,² the yen returned. My wife drilled me in the code without ever learning it herself. I bought the first commercially-manufactured amateur transmitter I had ever seen or heard of: an Eico Model 720, for \$25.

"What will they think of

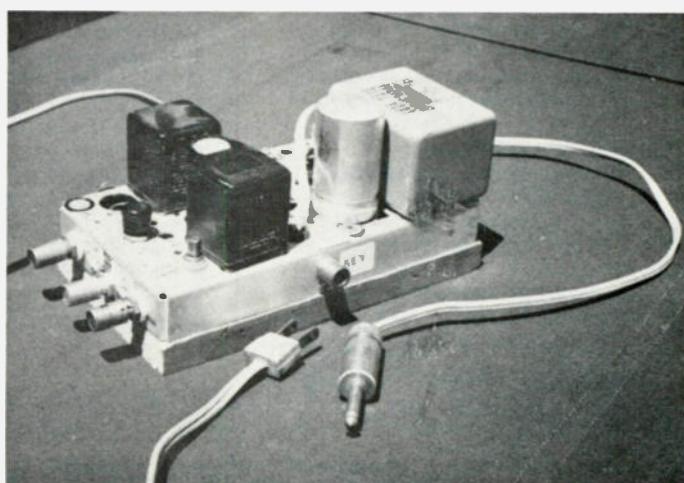
next?" I asked myself. "Somebody is saving us new amateurs a lot of trouble by building our transmitters for us."

That little jewel didn't work and it didn't have a manual or a schematic, and I didn't have a license yet, but I went right out and bought a nice brass key and a pair of coax connectors. A few days later I laid out thirty-five dollars for an Army surplus BC-348-Q receiver at a local swap meet. (Not a hamfest. I hadn't even heard of such a thing.)

WA7UKB was my "Elmer," my encourager,



This is the finished QSK unit, showing antenna jacks, ugly holes, and wooden base.



The finished QSK unit from its good side, showing mute adjustment, mute-defeat switch (the push-button), and mute and sidetone binding posts.

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TRANSMIT!!

YEEOOUUP YIP YAARRK
YIP - WOOOP YEAAAP
YIP WAAAHP...

"Mercy sakes, good buddy! My nice new receiver has gone to the happy hunting ground." It hadn't. When I threw the antenna switch back to "receive," the WWII surplus receiver made those beautiful sea-double-you sounds again.

I asked a lot of questions and listened to a lot of answers and bought a book about amateur radio and studied it. I built an electronic transmit-receive switch right inside the transmitter and thought I was the only ham in the world smart enough to do this, seeing that my cookbook-type instruction sheet showed it on a separate chassis with a separate power supply.

I bought another receiver. It was bigger and prettier than the BC-348, but it had no markings at all, so I could never tell anyone just what I had. Anyway, with this handy combination I could transmit by pushing the key down and listen by letting it up again. The receiver overloaded very nicely and merely woofed and grunted when I transmitted, and I actually made a few contacts—some of them out-of-state.

The next summer I found out about hamfests, and I went to the one at Flagstaff. I bought a Johnson Viking Ranger because it had a built-in vfo. This transmitter didn't work, either, but it had a manual and schematic. In fact, it had two manuals—both for the Ranger II. I took out the modulator and built in a T-R switch just like the other T-R switch. It didn't work quite like the other. In fact, it didn't work at all. I built two or three different kinds, following magazine articles and friendly advice. They didn't work either.

Back to the old knife switch. Back to turning the receiver gain way down to transmit and turning it back up again to receive. Back to lost QSOs.

Back to reading about ham instead of doing ham. I bought a big pile of 73s and other amateur magazines second-hand and spent a lot of time reading them.³ There are a lot of articles in the back issues that are worth reading. Some of them tell how to change from transmit to receive without having to throw half a dozen switches. Some of them tell how to do it without using noisy, old-fashioned relays. I laid out about a day's wages for PNP transistors in blister packs and proceeded to gather the other components. The day before the transistors came in the mail, I finally found what I was looking for all along:

How to do it with those noisy, old-fashioned relays.

Now, I had never lost the love of surprises ever since that hundred-pound grab-box twenty years before, and sometime along then I started buying two-ton grab-piles from Department of Defense Surplus Sales.⁴ Therefore, I had a very respectable junk box. It's about four feet deep in

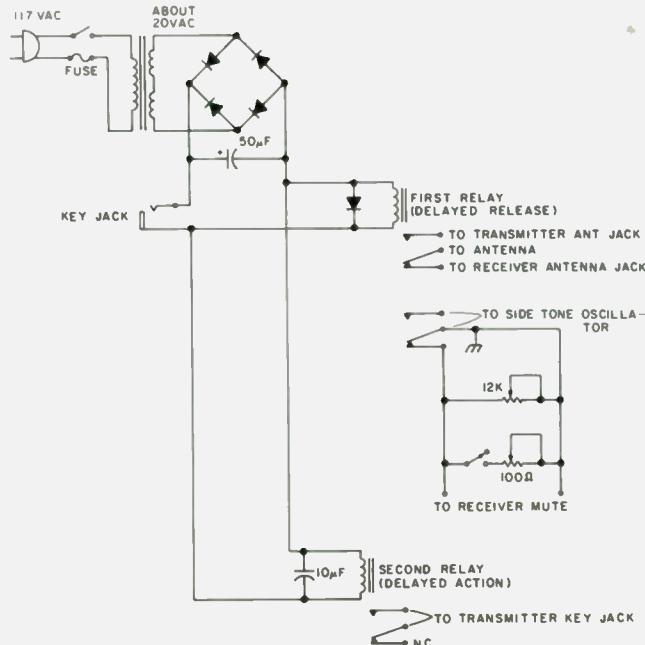


Fig. 1. Schematic diagram of break-in unit. The diode across the first relay delays its release; the capacitor across the second relay delays its action. Parts values are not critical. Contacts are shown with relays not energized.

not-so-neat rows all across my back yard. There are lots and lots of noisy, old-fashioned relays in MY junk box.

It seems the problem with relays is that the contacts cannot handle nearly as much current when they are opening and closing as when they are already closed. Sparks fly. Metal melts, or even vaporizes. The contacts get rough, pitted, oxidized, non-conductive. They may even weld themselves together. But if the current does not flow except while the contacts stay closed, those nasty things don't happen. (Unless, of course, you have a lot of current flowing.)

If you key the energizing coil of a DPDT relay and have one set of contacts key the transmitter and the other set change the antenna over from the receiver to the transmitter, pretty soon you fry the antenna change-over contacts because the finals are still putting out when the antenna connection opens at the end of each dit or dah. Sparks fly, etc.

You really need two re-

lays set to operate a very small fraction of a second apart. When you close the key, the first relay should change the antenna from receive to transmit and then the second relay should key the transmitter. When you open the key, the second relay should unkey the transmitter and then the first relay should change the antenna from transmit to receive. This way, there is no rf flowing in the relay contacts while they are opening and closing.

Here is the precious secret the Guru up on the high mountain peak in Northern India imparted to me, and now I disclose it to the world. (Actually, it has been disclosed to the world many times before, in amateur radio magazines, engineering texts, electronics magazines, the *Toymakers Journal*, handbooks, letters to the editor, over the air, and under the counter.)

You can delay the opening of a relay by putting a diode across the coil so that the inductive 'kick' when the current is shut off has a place to go; it goes

through the diode and back around the coil, and around and around it goes until it gets tired of going around and just lies down and quits due to the resistance of the circuit. This holds the relay closed a long time—maybe a thousandth of a second. When it finally, at long last, opens, a millisecond or so later, one hopes that there is no current left at the contacts to burn them up and weld them together. If your relay opens too slowly to suit your purpose, put in a few Ohms of resistance.

You can delay the closing of a relay by putting a capacitor across the energizing coil. When the current comes to close the relay, it has to charge the capacitor to a certain extent before it can close the relay. The delay is controlled by the capacity. This delay lets the other relay close and change the antenna over to "transmit" before the transmitter is even keyed. (See Fig. 1.)

This is not a secret, not a new breakthrough. But it was new to me. And very interesting. I went looking through my rows of surplus electronic scrap.

There was an aluminum chassis, about the right size, and it took me only half an evening to get the government parts off it so that I could put my parts on it. There were the relays, of course, and a filament transformer with enough winding to put in series to get enough volts. There was a solid-state rectifier bridge, a diode, and lotsa connectors. No two parts came off any one chassis in my junkyard; oops, pardon me. I mean junk box. Never bought a thing for this project.

Well, yes, it looks kinda funny, and there are holes in the chassis with no parts in them and circuit labels for circuits that aren't there any more. The labels

that count are just stick-ons, and there's that big capacitor that's merely taped onto the side of a relay. There's no panel, and wires stick out of every side except the bottom. But it works. And I did it my way. No one else ever had a QSK unit just like this one.

You're smart. You're careful. You can make a better, prettier one.

About those noisy, old-fashioned relays: They clack, clack, clack. Transistors don't. When my super-stable antique transmitter vfo drifts so far that I cannot hear my own sending on my super-stable antique receiver, I can still hear my sending, after a limited fashion, in the clacking of the relays.

These good government relays have extra sets of contacts and they are useful, too. Hook up one set as a mute and sidetone control. (See Fig. 1.) This thing worked the first time around. (That should prove there is nothing critical about it.) The receiver was silent when the transmitter was keyed, and it came back to life instantly when the key was raised. I actually made a contact or two that way before I tore into it to make improvements.

In order to hear the signal on the receiver, it cannot be altogether muted. (That's because I never got around to building that sidetone oscillator.) The diagram shows a variable resistor across the receiver-mute contacts of the first relay.⁵ It is there to adjust the loudness of the transmitter as heard in the receiver. Now I could hear myself send, but could not zero-beat the transmitter to the incoming signal without sending out a full-strength signal. It is a "no-no" to twiddle the vfo dial around with the finals turned on. It aggravates a lot of people to hear that

horrid WAH-YOOP sound when they are copying code, especially when they're copying the DX ham's callsign. The last thing I've done (so far) to this thingie is to add the "spot" switch and a second variable resistor across the receiver-mute contacts of the first relay.⁵ By closing that switch, I can hear the transmitter vfo and zero-beat the guy who's calling CQ before turning on the finals. By opening it, I can hear myself send without ripping my speaker cone and my eardrums. And the relays take care of changing the antenna from the transmitter to the receiver and back and turning the gain up and down as fast as I can key.

No. I haven't achieved Nirvana. Pardon me, I mean perfect break-in. But I have had many relaxed, satisfactory communications since I built this thing. Now that I write this, it occurs to me

that it is too complicated, too ugly . . . maybe a single relay would do—an open frame relay where I could get at the contacts and just bend them a little so that one contact makes before another. Let's see, a triple-pole, double-throw relay . . . and build it right into the transmitter . . . quiet it with sponge rubber, maybe . . . hmmmm . . .

I can see now that this is not the final chapter of "Quest." In fact, it is all prologue. ■

Footnotes

1. *Book of Knowledge*, 1957, Vol. 14, p. 5183.

2. The United States Army, Department of Defense, Washington, D.C.

3. Lockheed Amateur Radio Club, 2814 Empire Ave., Burbank CA 91504.

4. DoD Surplus Sales, Box 1370, Battle Creek MI 49016.

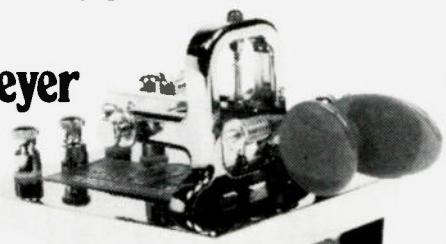
5. These values were found by experiment, and are good only for my particular combination of vintage equipment.

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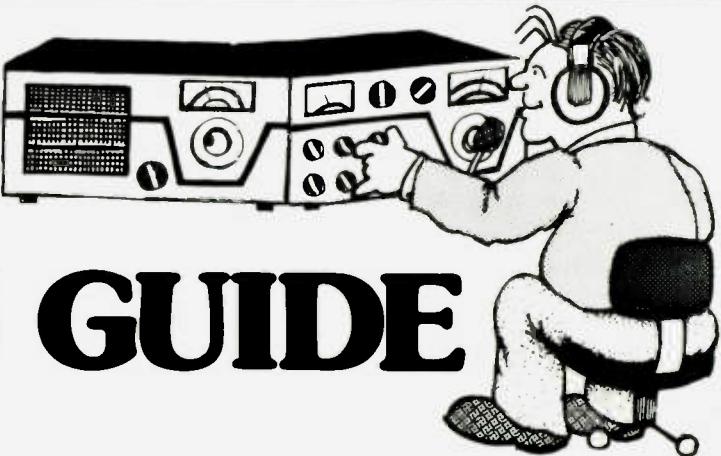
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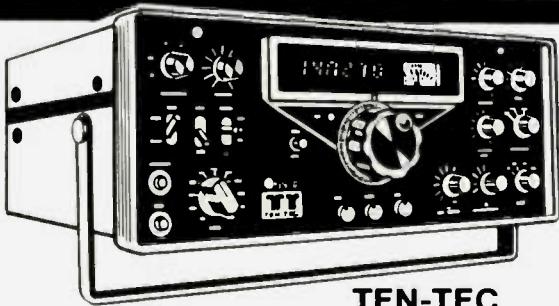


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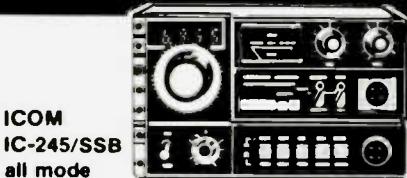


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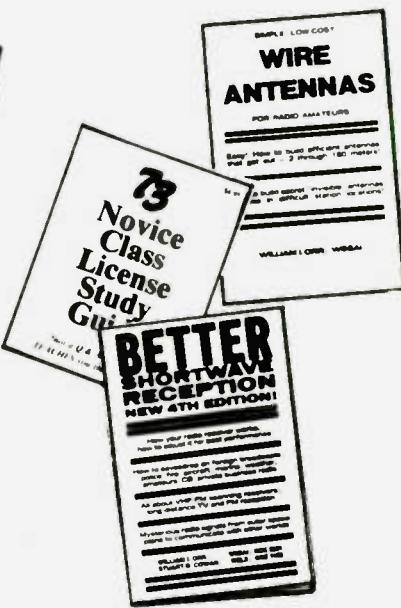
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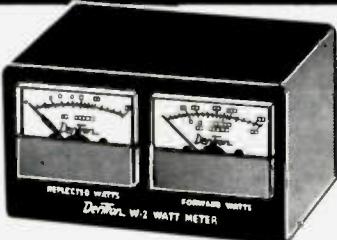
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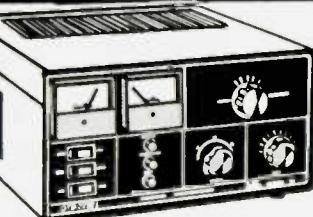
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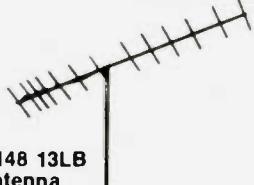
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The Amazing Audio Elixir

—this limiting amp is a cure-all

Back in the days of the Old West, there were groups of traveling salesmen who sold snake oil and other cure-alls. The claims made by these people and their medicine shows were that if you bought this wonder elixir, it would cure anything that ailed you. Here is a modern-day electronic circuit which will "cure all that ails you." The circuit presented here is a general-purpose amplifier which can be used to solve many problems and increase the performance of many audio circuits. The unique feature of this amplifier is that it will take an input of 100 mV to 10 V while the output level stays the same. The output is ad-

justable from no output up to about 8 volts peak-to-peak. You are, no doubt, asking what it is used for. Let me describe a few uses which will get your mind working and then you can take it from there.

In a repeater system, a tone decoder is used for control. The tone decoder works best with some precise input level, and the users are always tweaking the pad levels to make their pads work the system reliably: Drop an audio limiting amplifier between the receiver and the tone decoder and the problem is gone. Another use is in an autopatch system. Put one of these little units be-

tween the phone line and the transmitter and you will always be able to hear the person on the phone at the normal level.

For you computer buffs, take this unit and put it between the output of the tape recorder and the tone input to your interface unit and get constant levels even as the tapes age. My 2-meter SSB rig does not have agc in it. When I have the volume high enough to hear a DX station and a local comes on, the level change inverts my speaker cone and my chair (with me in it). I lifted the wire from the top of the volume control and placed one of these units between it and the volume control. Now all the stations come out at the same level, at whatever level the volume control is set, and there is no more ear damage for me.

I know of one person who set up a room microphone and used this unit to record anyone talking in the room without moving the microphone or having to ride the tape recorder's input control. With touch-tone™-type signals, place a group splitting filter in front of two of these units and independently level

each tone group to remove any twist or imbalance in the tones. How about a speech processor? Use this unit in a fast-agc mode and drive the input to your SSB rig at full level all the time. Another use might be as a "linear clipper" in a modulator. Drive the unit below the point where the output limits, and if the signal through it gets too high, the limiter cuts back and won't allow excess audio to pass. The advantage of this unit over normal clipping methods is that this unit does not clip, and thus it does not add clipping distortion. Enough of these ideas. Let's go on to see how it works.

How It Works

See Fig. 1. U1 is an op amp and is simply a gain block. The gain of U1 is set by the values of R4 and R6. As shown, the amplifier is set for a gain of about 200. The output is sampled by C3, CR1, CR2, and C4. This network produces a dc voltage directly related to the ac (audio) output of the amplifier. This dc level is fed to the gate of FET Q1, which is an inexpensive Motorola MPF111. This FET is not critical and any

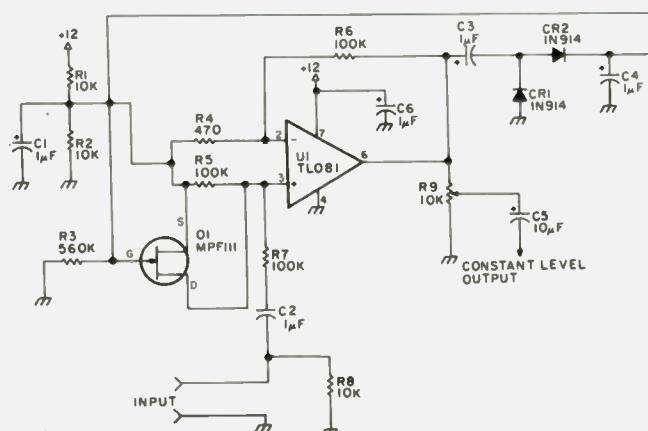


Fig. 1. Audio-limiting amplifier diagram. (This diagram is supplied with printed circuit boards.)

equivalent FET will do. R1 and R2 form a voltage divider which is used to bias the FET source and the op amp inputs at a point exactly one-half way between the power supply voltage and ground. C1 ensures that the bias point is at ac ground. With no detected voltage from C4 to the gate of Q1, the gate of Q1 is at ground, which is negative to the source, and the FET is in a "pinched-off," or nonconducting, condition.

The incoming signal comes in through R7 and is amplified by U1. The output is detected and the dc voltage is fed back, which lifts the gate of Q1 from ground and Q1 starts to conduct. As Q1 conducts more and more, it shunts the input signal away from the op amp input. R7 and R5 act as a voltage divider and, with the FET, R5 acts as a variable resistor. The combination acts as an elec-

tronically-operated pot, with the "wiper" connected to the input of the amplifier. Once the detected output reaches a certain level, it will cut back the input signal and not allow the output to go any higher.

The purpose of C2 is to block the bias voltage from going out the input. R8 is used only to terminate the input and should be a value that the driving circuit wants to see. It can be any value because it has no effect on the operation of the circuit. Since U1 is a closed-loop amplifier with a fixed gain, the output will drive anything as long as the output capabilities of the op amp are not exceeded. Even though there is some circuit loading, I have even driven small speakers with this unit and obtained output volume which was usable for testing purposes. The amplifier is protected so you do not

need to worry about damage.

This unit is designed for 12-volt operation, which is near its lower operating voltage limits. The voltage is not critical, and as long as the voltage limits of the capacitors are not exceeded, it will work over a wide range of voltage. The lower limit is 12 volts and the upper limit is greater than 30 volts. If different agc release times are desired, the value of R3 may be changed. The larger the value of R3, the longer, or

slower, the recovery time of the amplifier. When used as an agc, it has a fast attack time and a fast decay with the values shown.

In conclusion, this article was meant to be mostly an idea article and describes a building block which I have found to be very useful. Circuit boards may be obtained from the author for \$3.50 each and are top-quality glass-filled, drilled boards with silk-screened parts layouts on the surface. ■

Parts List

Designation	Description	Qty.
U1	TL081 BiFET op amp	1
Q1	MPF111 FET	1
CR1, CR2	1N914 signal diode	2
R4	470-Ohm, 1/4-Watt, 5% resistor	1
R1, R2, R8	10k, 1/4-Watt, 5% resistor	3
R5, R6, R7	100k, 1/4-Watt, 5% resistor	3
R3	560k, 1/4-Watt, 5% resistor	1
R9	10k pot, multi-turn variable resistor	1
C2	.1-uF mylar™ capacitor	1
C1, C3, C4, C6	1-uF electrolytic capacitor	4
C5	10-uF electrolytic capacitor	1
	PCB	1

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Are your bored with rotary switches in your projects? Even if by some slim chance you are not, here are a few simple remedies for replacing those tiresome rotaries with a sequentially-stepping counter that lights one of a string of LEDs each time a push-button momentary switch is pushed.

The cost of such a system is about as much as

you would pay for a new rotary switch, but the main attraction is that it makes that "electronic wombat finder" of yours a real eye-catcher that is fun to use.

The devices described here are primarily CMOS, because I am a devoted CMOS fanatic. With proper attention to the pertinent peculiarities, however, any other logic family will work well.

A two-position, two-deck rotary switch eliminator is shown in Fig. 1. The switch portion of the circuit is a set-reset flip-flop that gives clean, bounceless low-to-

high transitions to clock the counter. The counter is a 4013 D-type flip-flop wired as a toggle. It divides the incoming clock frequency by two and transfers data from the D input to the Q output on the positive edge of the clock pulse. Each push of the button puts a low to high signal at the clock input. With CMOS, these levels are essentially ground and the supply voltage. We will call these low and high levels "0" and "1" for clarity.

The characteristics of a D-type flip-flop wired as a

toggle, with the Q_1 output fed back into the D input, is that for every 0 to 1 transition on the clock input, the Q output changes, and then remains at that value until the next positive going clock pulse comes along. This means that it has two stable conditions. If the Q output was at a 1 level, then a single push of the button would change it to a 0. The next push would send it back to a 1. The \bar{Q} output does the same, but, of course, is of the opposite polarity.

As each of these outputs assumes a 1 level, it will turn on its LED driver, the 75492. The 75492, which has six drivers per package, is a power amplifier to supply the current requirements of the LED. The LEDs current is limited by a resistor whose value will be determined by the supply voltage and the desired current. A reasonable formula for such calculations is given in Fig. 1.

Speaking of power, CMOS works fine on 3-15 volts, but the 75492, a bipolar device with CMOS compatible inputs, is limited to 10 volts. Anyway, whichever one of the outputs, Q or \bar{Q} , is at a 1 level will cause the 75492 driver

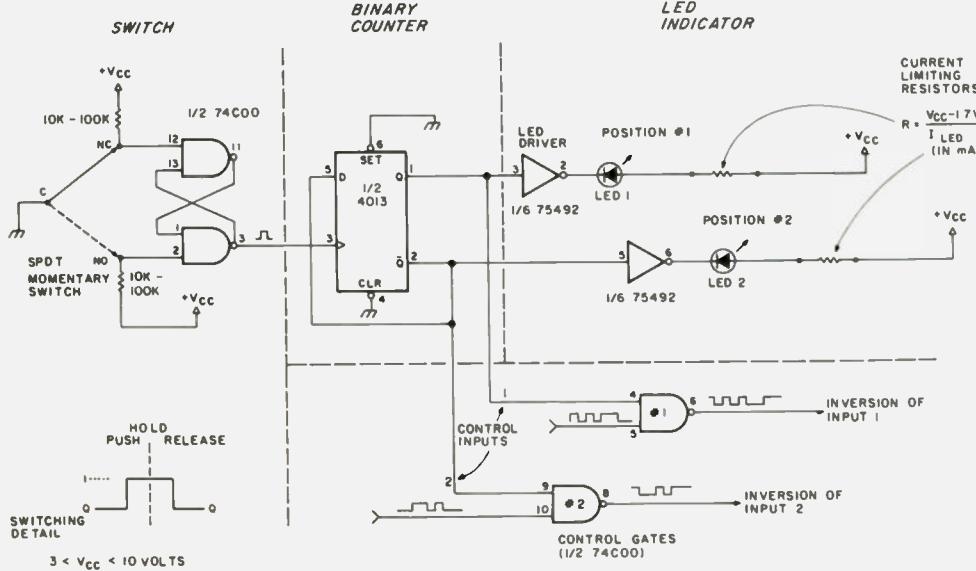


Fig. 1. Two-position rotary switch eliminator.

to go low, which will forward bias the LED, turning it on.

So much for the indicator, but what will it indicate? We need our switch to control something. This is accomplished by the remaining 74C00 2-input NAND gates labeled Control Gates. This section of the circuit is the second "deck" of our two-deck switch eliminator.

A 1 from either the Q or Q̄ flip-flop at one input of the 74C00 will allow an inverted replica of a signal at the other input to pass through the device. A 0 at the control input will keep the output at a 1 level regardless of the condition at the other input. As the button is pushed again, the control is shifted to Control Gate 2 and Control Gate 1 is disabled, remaining at a 1 level.

Because of the binary nature of toggle flip-flops, any number of switch positions that is a power of two can be readily made by cascading these binary dividers until the desired number of stages is reached.

A four-position rotary switch eliminator is shown in Fig. 2. With these two flip-flops come a possible four combinations of the Q outputs, Q_A and Q_B . They are 0,0; 1,0; 0,1; 1,1. The next sequences are only repetitions of the first. The sequence is as follows: Assume both Q_A and Q_B are at a 0 level. After the button is pushed once, Q_A goes to a 1, but Q_B is still at a 0 level. So far the count sequence is 0,0; 1,0. The next push gives a 0 at Q_A , but now Q_B goes to a 1. The sequence is now 0,0; 1,0; 0,1. The third push of the button flips Q_A to a 1, but because Q_B changes only half as often, it remains a 1. The sequence is now complete: 0,0; 1,0; 0,1; 1,1. The next push would cause Q_A and Q_B to become 0,1.

Now we have to decode

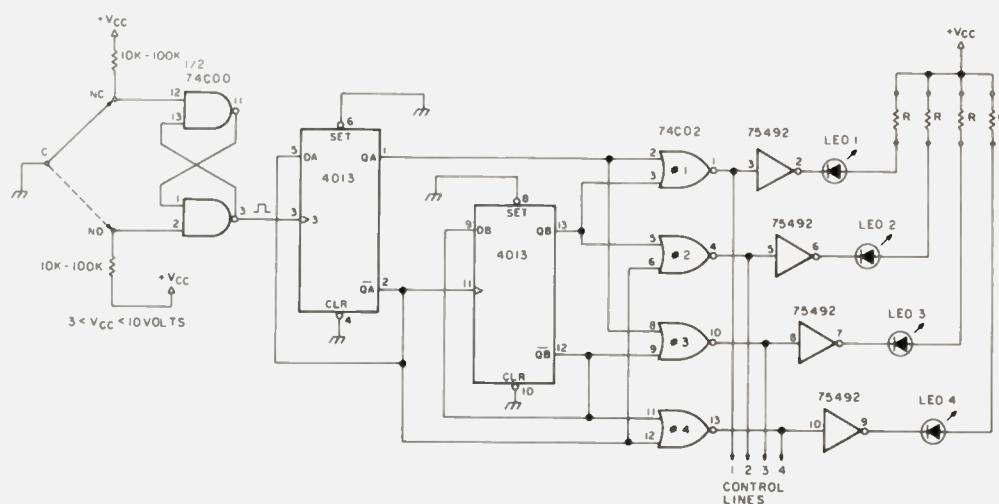


Fig. 2. Four-position rotary switch eliminator.

each of these pairs of outputs to light four LEDs in sequence, one at a time. The 74C02 quad 2-input NOR gates do this trick neatly.

The NOR gate is a great "zero detector" because it will give a 1 output only when all of its inputs are at a 0 level. The NOR gate fed directly with Q_A and Q_B would output a 1 level only when the count of 0,0 was reached, turning on the 75492 LED driver, illuminating LED 1. The next state in the count sequence is 1,0. This will extinguish LED 1 because the 74C02 no longer has the conditions required to turn on LED 1.

For LED 2 to turn on, we can use Q_B directly because it is a 0 level. Instead of Q_A , though, we'll use Q_A , which is a 0 when Q_A is a 1.

Now 74C02 #2 has the proper zeros on the inputs to give a 1 output to turn on LED #2. The next state of the counter results in a Q_A of 0 and a Q_B equal to a 1 level. 74C02 #3 will be activated by using Q_A and Q_B , the exact opposite from the requirements for 74C02 #2.

The final push gives a 1,1 condition, which needs both Q_A and Q_B to be used as 0 inputs to 74C02 #4 to turn on LED #4.

So much for a four-position switch, but how about an odd number like three? What we'll do is to start with a four-position counter and use the fourth state of 1,1 to reset the circuit back to 0,0. The counter will think it is still a four-position counter, and only we will know the difference.

The three-position rotary switch eliminator is shown in Fig. 3. The significant difference between it and the four-position counter in Fig. 2 is that the 74C02 #4, which detects and decodes the 1,1 condition, is used to reset both halves of the 4013 to zero, giving a count sequence of 0,0; 1,0; 0,1, and back to 0,0.

The point of this decoding business is to ensure that each and every combination of counter outputs results in only one LED on at a time. Otherwise, we will have the equivalent of a rotary switch with shorted contacts.

If you are interested in readable, useful information on counters, try one or more of Don Lancaster's Cookbooks. The "reci-

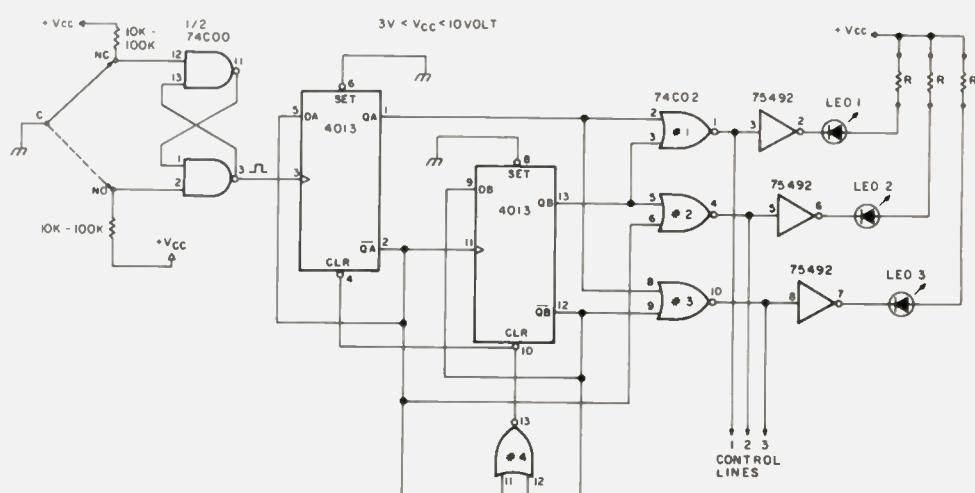


Fig. 3. Three-position rotary switch eliminator.

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stuffing.²
Simulating any number

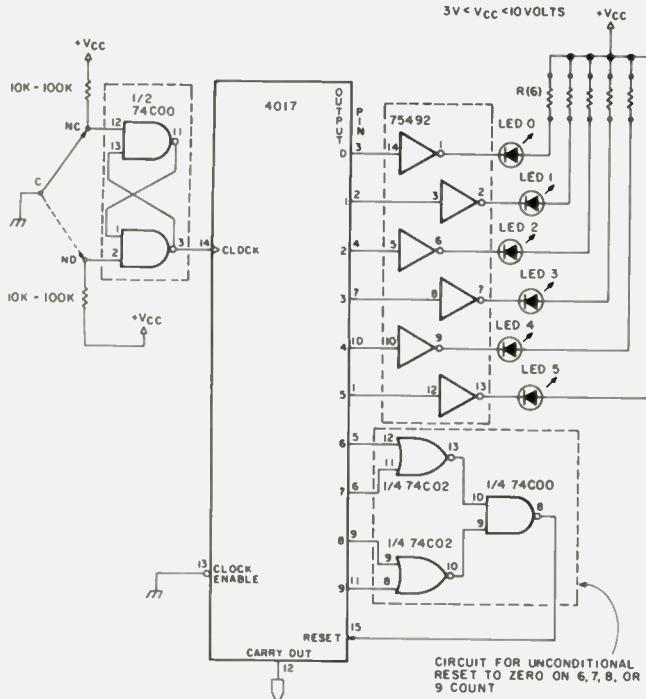


Fig. 4. Six-position rotary switch eliminator using a 4017 decade counter.

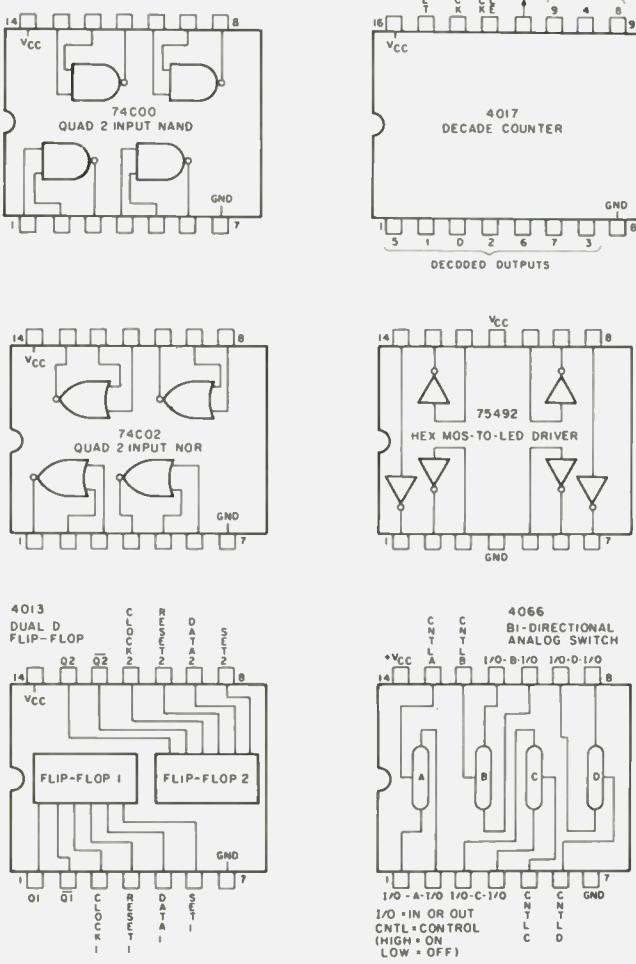


Fig. 5. Pinouts of the ICs used (top view).

of switch positions is possible by remembering two basic guides. One is to be certain of the sequence of the outputs of the flip-flops as you push the button. This knowledge will allow you to properly decode the outputs to control the LED lighting and whatever else you have in mind. Equally important is to be sure that all the output combinations are accounted for, either for proper decoding, or for resetting the counters, as shown in the three-position counter. Any combination you miss will at best give the effect of a dead switch position, or, as previously mentioned, a shorted switch.³

An easy way to avoid concerning yourself with much of the above trouble is to use a resettable counter like a 4017. Any number of switch positions from 1 to 10 can directly be implemented. The 4017 is a five-stage divide-by-ten Johnson counter with ten unique decoded outputs. Johnson must have been clairvoyant, because this sixteen-pin package is just what we need.

The example in Fig. 4 is a six-position rotary switch eliminator. This circuit will count from 0 to 5, then start over again with successive pushes of the button. The two 74C02 gates and the 74C00 gates eliminate any hazard of unwanted outputs, in this case, a 6, 7, 8, or 9 count. You'll recall that the NOR gate gives a 1 output if all inputs are 0. Well, any active 6, 7, 8, or 9 output will give a 0 at the output of the one of the NORs, and this 0 will drive the 74C00 NAND's output to a 1 level, which will just happen to reset the counter to 0. Now at any count from 0 to 5 the 6, 7, 8, and 9 outputs are all 0. Following these through the NOR and NAND logic gives a 0 on the reset line, which is just what we want, that is, no reset until a 6, 7, 8, or 9 appears. The 0

through 5 outputs control the lighting of their respective LEDs through the six gates of 75492 package. These lines can also be used to control other functions, as in Fig. 1. Easy enough. The 4017 also has a carry out, which allows cascading 4017s together for long count sequences. More components, a little imagination and a calloused index finger are the only requirements.

So far, only digital controls have been discussed. By using 4066 bi-directional analog transmission gates, low level analog (or digital) signals up to 40 MHz can be faithfully controlled with minute distortion. Just substitute 4066s for the 74C00 Control Gates as in Fig. 1. Pinouts for all the devices mentioned are shown for reference in Fig. 5.^{4,5}

As an extra attraction, these circuits also have built-in service aids. The LED indicators can act as diagnostic pointers by "illuminating" your logic and/or wiring errors. With all this in mind, "switch" to these rotary eliminators for your current and future projects. If you have the capacity, I'm sure you'll get a charge from designing and building them. You, too, can be the envy of friends and rivals alike! If I have etched them into your memory, then they will certainly be difficult to resist. ■

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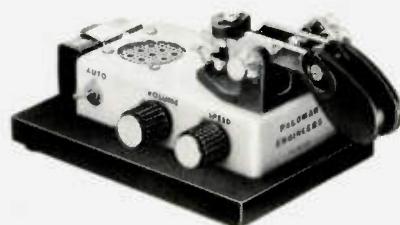
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Hans Peter
c/o 73 Magazine

In August of 1975, while I was a long-term patient in the hospital, I received an AM mobile transceiver. This was before the FCC changed the Part 95 regulations to allow idle chitchat one month later. I did not receive my FCC call letters until October, two months later, but of course I transmitted before then.

Roll Your Own

On all crystal-synthesized 23-channel CB rigs, there was a blank spot between channels 22 and 23. It was normally an inoperable channel that, I thought at the time, was the position to use when using the PA function. After meeting a few other CBers and discussing equipment, I learned that on a few rigs, by setting the dial between the blank spot and channel 23, it was possible to "roll in" a non-CB channel, commonly called 22A. There was also another channel called 22B, but this channel could

only be installed by adding at least one crystal to the rig. No one really knew what these channels were for, but supposedly they were business channels.

I could not roll this channel in on my particular rig, so I wired in 2 switches to connect where the rotary switch was not making contact. It worked, and I was very excited to discover this modification. It seemed that most people thought that if you could not roll it in, then it was not possible to get this channel. At that time it was a quiet channel and at first only a few "elite CBers" had it.

I did read in an advertisement for an 11 meter vfo that 22A and B were for business traffic, but I never did hear any, and I could not find anything about this in FCC rules and regulations.

I also started modifying other CBers' radios to get this channel. It could be done by cutting a ground wire, or adding a jumper wire, or installing a toggle switch if the manufacturer

had gone to great lengths to internally defeat this channel. I charged five dollars to do this modification, although I knew of someone else with a First Class license who charged twenty dollars.

When people would get on frequency, they generally would say "break A" or "break 22A." Now this frequency has become channel 24. If you hear someone on the air saying "break A," they probably are using a 23-channel transceiver.

I met another fellow on 22A who was also interested in the electronic aspect of CB, and we became good friends. We both learned as much as we could about radio modification. We also did radio work for other CBers.

Moving Up

I eventually moved on to bigger and better things, namely getting into the HF aspect of 11 meters.

I studied my crystal synthesizer and eventually figured out that by changing two master crystals I

would get 23 channels just above the class D channels. At first I had no idea where to get these crystals made. Finally, in a CB catalogue, I saw I could order crystal certificates and have the crystals made. I thought perhaps the company I sent the certificates to would not grind the crystals for me because they would result in out-of-band operations, but this did not happen.

After about 3 months of waiting, the crystals arrived. I thought that by just unplugging the original master crystals and plugging in the new ones, I would be all set. However, only the transmitter would work. The receiver was dead. I eventually found out I had to retune the synthesizer cans for each set of master crystals. But it did work and I got HF capabilities. Later, I tuned the receiver for a happy medium so both sets of crystals worked, but the receiver performance suffered and I got all kinds of receiver images.

CB shops eventually

started selling HF synthesizer crystals. These would give 4 or 5 new channels per crystal. They were hot items and sold out quickly. Very few CBers even knew how to wire in these crystals, and here I was of service. The crystals sold for five dollars. I would install a 2-pole toggle switch with one empty socket and an HF crystal for twenty dollars.

Gadgets

CBers are almost helpless when it comes to doing even simple modifications or repairs. Even when a microphone cord needs to be resoldered or a CBer buys a preamplified mike, he has to pay someone to fix or install it. This lack of technical knowledge leaves CBers open to many rip-off artists. I have seen standing-wave meters advertised as 40-channel standing-wave meters, as if the ones manufactured during the days of the 23-channel sets were inferior.

I have seen an item called a crystal vfo. What is it? It's a little box with a twelve-position rotary switch connected to twelve crystal sockets with a trimmer cap for each socket. Theoretically, one of these could give a crystal-synthesized transceiver 60 new channels. All this for only \$100, less the crystals.

Other items in demand are tweety birds and pingers. The pinger produces a short feedback-like squeal of short duration when the microphone is keyed. The Browning Golden Eagle had this built in. Because of pressure from the FCC, the new 40-channel Brownings do not have this feature. It can easily be added because the pinger is still in the radio but is not connected when the set comes from the factory. Pingers are now sold as add-on units by

an independent firm for about \$18. These can be attached to any radio.

The tweety bird produces a chirping sound. They are often used to attract attention, but can cause severe bleed-over on adjacent channels. Prices for tweety birds are around \$15 to \$20. There are full page ads for both of these noisemakers in CB magazines, even though they are prohibited by Part 95 of the FCC rules and regulations.

Regular tunable 11 meter vfos are made which plug into one of the crystal sockets in the transceiver's synthesizer. There are two very popular models.

One, made by Siltronix, has a simple two-transistor LC circuit and is inclined to drift unless the unit is left on continuously. Few CBers know that these can be adapted to any transceiver by adding or subtracting some capacitance to or from the LC circuit. The other popular model is made by PAL. These are rather complex and use a crystal-controlled reference oscillator.

I have seen used units at hamfests going for \$175 to \$200, which is the same as the retail price. The funny part about eleven meter vfos is that in advertisements it specifically states that it is illegal to transmit with vfos, as if that will stop anyone. I charged \$20 to install a vfo. I arranged it so the vfo could be disconnected from the transceiver, for obvious reasons.

The clarifier on a single sideband set could be modified into a quasi-vfo by changing some components or readjusting trimmer resistors. This increased the sliding range from ± 500 Hz to about plus 2 kHz and minus about 13 or 14 kHz. Now the clarifier could slide into RC channels which are between class D channels.

In the world of CB, if something is illegal, it is even more desirable to have, and the CB consumer will pay an arm and a leg to have it.

Having a transmitter "peaked up" is another service in demand. Rarely can more than 1 extra Watt be squeezed out of an average CB transceiver and still have just a touch of upward modulation. Overdesign is not common. But a CBer usually thinks, "It may help me get out an extra mile." I usually charged \$5 to peak the transmitter, but a licensed repairman could get \$20 to \$25. Of course, he had to do this under the counter. The big money in CB is in modification, not repair.

HF Lingo

How do HFers refer to illegal channels? They think of the frequency in megahertz and use the first three digits after the decimal point. So 27.415 would be just 415. This is the first channel above channel 40. From channel 26 through 40, the frequency can be determined by just adding a 5 to the end of these digits. Channel 40 is actually 27.405, channel 34 is 27.345. By extrapolation then, 27.505 would be channel 50. Many AM HFers use this method of nomenclature, and you will hear "Break channel 43" on 27.435.

Eleven meter single sideband is a rather new and very different development. Almost all CB sideband is on the lower side. Upper-side users are looked down upon and sometimes even harassed when on the air.

Sideband conversations are formalized. They are very similar to ham QSOs, which they are intended to mimic. No ten codes are used as on AM—only Q codes are used. There are no AM handles either; everyone uses first names.

If you are an AMer new to sideband and have picked up the habit of saying 10-4, you will often be politely asked to try not to say it. "Roger" replaces 10-4 on sideband. There is very little of the stereotyped channel 19 lingo on sideband. But when skip starts, it is funny to hear sidebanders call out "CQ DX" over and over.

To replace handles, sideband operators have club call letters. The first CB sideband clubs formed used the number designating their state's place in the order of all states' admittance to the Union, followed by a W, which stands for World. For example, California would be a 31W state because it was the thirty-first state to be admitted to the Union. Newer clubs are using an N prefix, for National, so Georgia would be a 5N state. Other abbreviations include A for America and X for X-ray. It seems like the whole alphabet is being used.

An AM operator will generally get mad when someone else uses the same handle. Sidebanders will also get disturbed when someone else has the same club call letters. How can this happen? Well, 36A3536 could be the 36 Alpha or the 36 America club (Nevada) when written down on paper.

Many sidebanders get a number from every club in the area so they can use call letters to match the person on the other end of the QSO. The use of club letters makes a neater and more efficient operation on the crowded Citizens Band.

When the DX is in, it is easy to figure out what states other operators are in. Club call letters replace FCC call letters on the HF band. During local QSOs on class D frequencies, both FCC and club call letters are generally used.

An AMer might say, "This is the Hill Billy, KABC 1234, going 10-7 for the night."

A sidebander might say, "This is 36-National-1234, KXYZ 9876, QRT and QSY to bed."

Using Ham Gear

Most HF gear is modified CB gear with possibly a linear amplifier, but converted amateur gear is most popular among sidebanders. It may not be apparent, but amateur equipment has distinct advantages and actually makes sense. An amateur transceiver has a dynamite receiver. Its sensitivity and selectivity will beat any piece of CB equipment hands down.

Citizens Band is far more crowded than any ham band. In the evening, signals abound as people turn on their radios to have a conversation. Adjacent channel interference is unbelievable. Look around an average suburban area sometime, and see how many CB base station antennas are on the rooftops. Citizens Band has truly become radio communications for the people.

Before CB expanded to 40 channels, FCC type acceptance rules were very lax. Very inexpensive 23-channel transceivers were junk. Not only were they severe generators of TVI, but when a preamplified microphone was added, the transmitted signal was 15 or 20 kHz wide.

And think about this: The Cadillac of CB radios is the Browning Golden Eagle. This 40-channel set now retails for about \$1000. A Drake or Yaesu has a transmitter that can be tuned and has complete range over the 11 meter band after relatively simple modification. The price is also several hundred dollars less. The cost of a moderately priced CB transceiver, external vfo,

and linear amplifier is about the same as that of a piece of amateur gear.

Amateur transceivers put out a clean signal. There is very little spurious radiation and harmonic content. If everyone ran ham equipment on eleven meters, there would be virtually no TVI and far less adjacent channel interference.

There seems to be little interference to licensed users of the HF band, mainly because the business band is FM and usually 5 kHz off from most HF traffic. Business-band licenses are not being renewed because of expansion to 40 channels, and their signals are getting fewer all the time. There are a few business-band stations between the newly allocated CB channels, but few CBers realize what the buzz from the FM signal is. As far as government use of the HF band goes, it probably exists, but I have never heard any government transmissions.

Practically no one worries about being caught by the FCC. There may be a scare now and then that the FCC is in town, but I have yet to hear of someone I know being apprehended. With so many people into CB, it just cannot be controlled easily.

HFers are generally the CBers who go for ham tickets. After obtaining a ham ticket, they generally stop HFing or curtail their activity greatly. Getting a ham license fills their DX appetite and their HFing is done less and less after that. But most never leave CB for good because they still have too many friends there.

There are some CBers operating amateur gear who will not use a non-class D frequency. Most of these people are sidebanders. When CB had only 23 channels, only one of those channels came to be

known as a sideband frequency. This was channel 16. As the popularity of CB began to rise, so did the crowding. Occasionally, AMers strayed onto this frequency. Because of the hassle, many people held their conversations about 60 kHz or more above 23. After 40-channel expansion took place, the top 60 kHz of the present class D frequencies, channels 35 through 40, became unofficial sideband-only frequencies, mainly because they were beforehand. Because these frequencies were legalized, some sideband operators, including those running ham gear, quit operating out of band because they felt they now had a place and did not want to lose the new frequencies to AMers.

As a rough estimate, about $\frac{1}{3}$ to $\frac{1}{2}$ of the total CB population has probably at one time operated on illegal frequencies. Most of these operators do so when DX is in. When conditions are quiet, there is little non-Class D activity. In fact, there is more illegal AM activity when conditions are quiet. The sidebanders seem to stay put.

Special Activities

There are a few unusual activities that take place in sideband HFing. Slow scan television is around 27.985, just 15 kHz below 10 meters. I have personally never viewed an SSTV signal when the DX is running.

Round table conversations happen frequently. Three or more stations from different parts of North America link up and talk for an hour or more. Virtually all of these HFers use amateur gear and beam antennas.

Phone patches are popular. If an HFer QSOs with another whose location is near his relatives, he will dial up a phone patch without a second thought.

Perhaps one of the most unusual HF clubs is the Eleven M Club. To join, you must own and operate your own business and run all Collins equipment. The club's standby frequency is 27.765, and they only want their members to use this frequency. They generally appear only when DX is running. And the Eleven M is not what you think. The club was supposedly started by eleven millionaires. A rumor circulating about them is that most of these guys are hams who want something a little different to do.

There is also illegal activity below channel 1. For some unexplainable reason, it is all AM activity. When skip starts, this spectrum of frequencies gets filled with traffic. It extends about 300 kHz or more below channel 1. The frequency 26.800 is a popular one because it is exactly 455 kHz below channel 23. A receive crystal for channel 23 put in the transmit crystal socket is a common way to get this frequency.

CB's Popularity

Because CB became so popular in a short period of time, the FCC had to legalize idle chitchat. I really do not think it affected CB's growth. Psychiatrists have come up with many reasons why CB has become so popular so quickly, but I do not agree with any of these.

I think there is a little bit of a ham in everyone. Most CBers do not know how their radio works, but why is that so bad? A lot of people, including some hams, do not understand how a Polaroid camera works or understand every part of a car, but this does not prevent them from using these items and getting good results.

What attracts people to CB is the fact that they can push a button on a microphone and talk to someone

else across town or further away. All this can be done without interconnecting wires. This is what attracted me to radio and this is what initially attracts every ham to amateur radio.

CB has proven there is strength in numbers. As more license applications came in, the FCC relaxed the rules. They allowed higher antenna placement, permitted idle conversation, changed the rules so that only the sender needed to use his call letters, and expanded the number of channels. Most of these changes happened in a relatively short period of time. No one really worked together as a group to achieve these ends, although I will admit that the EIA lobbies wield quite a bit of force.

Yet 350,000 hams, who generally know radio well

and work in unison for the most part, cannot persuade the FCC to allow phone operations in a large segment of the 20 meter band now allocated to CW.

Many CBers would like even more expansion on 27 MHz, but the FCC will not expand any further because of intermodulated products resulting from 455 kHz i-fs. Now that 220 MHz is not available, CB will probably end up around 800 or 900 MHz if there is any further expansion. Antennas with huge amounts of gain would be possible, and so would moonbounce. It would be an ideal hobby band. Skip would be all but totally eliminated, and so would interference to other devices. Skip on 11 meters is the main reason HFing exists. HF activity just is not there unless skip is present. ■

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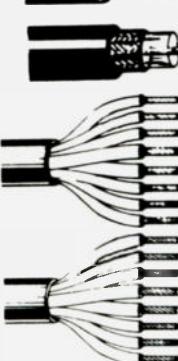
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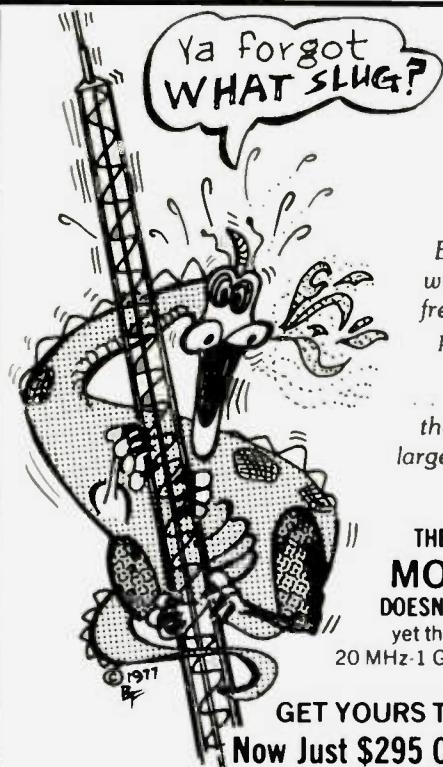
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The IC-502 is an excellent 6 meter transceiver for the money, but sometimes a little more power is very useful, es-

pecially during skip openings. Here is a linear amplifier that you can build at little or no cost, which runs about seventy Watts input. The heart of this amplifier is a surplus tube-type low-band Motorola FM transmitter strip. I bought mine for \$1.00 at a local hamfest. Only the final amplifier of the transmitter strip is actually used, so the condition of the

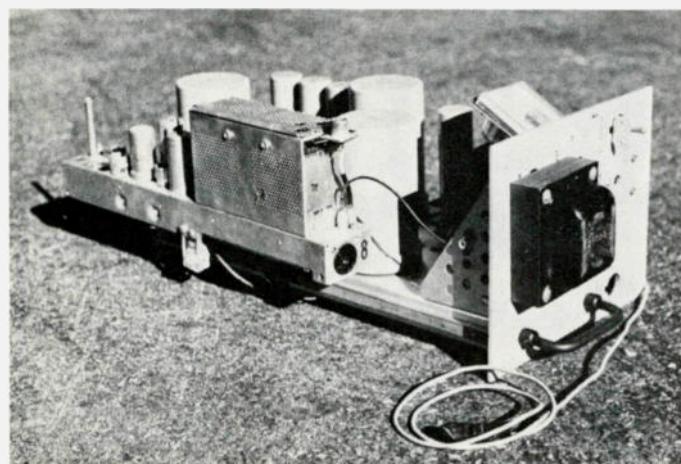
strip is unimportant. The advantage of my scheme is that the critical VHF wiring, neutralizing, and shielding, as well as the procurement of rf components is all taken care of for you. Essentially, this project consists of modifying the transmitter strip, building a power supply, and creating the relay switching for the amplifier.

This amplifier will also work well with low-power AM transceivers, such as

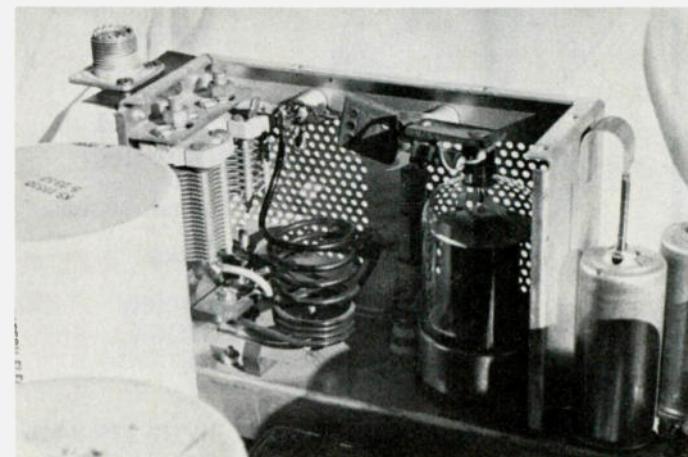
the Heathkit Sixer and the Lafayette HA-650.

Selecting a Transmitter Strip

This conversion is based on a Motorola TA104 series low-band transmitter strip. These can be easily recognized by their long, thin appearance, and a large multi-turn tank coil. This transmitter series covers the 25-to-54-MHz range in 5-MHz segments. To find the



Completed amplifier. The power transformer is in the foreground, and C1 can be seen attached below the transmitter strip.



Power amplifier coil. Note cut-down tank coil.

frequency range of a particular transmitter, first locate its identification numbers which are stamped in black ink, usually around the final amplifier compartment. These numbers should either be TU113 or TU113-C2. After the identification number, there is a one- or two-letter suffix. This denotes the frequency range of the transmitter. Before you go out shopping for your transmitter strip, make a copy of the suffix frequency code shown in Fig. 1's caption and keep it handy.

The unit covered by this article is an MM-type (35-40 MHz). This type often shows up at low prices because it requires a major conversion job to move it up to 6 meters or down to 10 meters. This poses no problem for its use in this project, since only the final amplifier is employed and only two coils need to be modified.

One final note: If your

transmitter strip is other than the HH (50-54) type, be sure to remove the T-section harmonic filter. This filter is located next to the antenna relay which is directly underneath the final amplifier compartment. It consists of two air-wound coils and three capacitors mounted on a tie lug strip. Remove this filter and run a wire directly from the output coupling coil, L2, to the antenna relay.

Construction

A heavy chassis should be used for the base. In my prototype, I used an old dc power supply as a starting point. The main power transformer is not critical. One from an old TV set will work fine. The bias transformer is a six-volt filament transformer connected in backwards. The high-voltage diodes are from Poly Paks and cost \$1.00 for five. Be sure to use diode-protection resistor-

capacitor networks across each diode. Also, be sure to use bleeder resistors across the filter capacitors. The plate milliamp meter is noncritical, but be sure that it is well insulated above ground.

To convert the transmitter final amplifier, it is necessary to use a grid-dip meter. If you don't own one, check with your local repeater group. You will probably find a ham who will loan you one.

To modify the transmitter strip, first cut the tank coil down four turns. Check for resonance at 50 megahertz by rotating the plate tune control. The resonance dip should occur with the plate tune capacitor half meshed. Next, remove T3, and take off seven turns from the top winding. With T3's top slug halfway in, connect capacitor C1 in parallel with the top winding of T3, and check for resonance at 50 megahertz. I used a vari-

able capacitor for C1 because T3's slug is hard to tune without a special tuning tool. Now replace T3, and complete the wiring. I mounted C1 on a piece of scrap plastic, but any suitable insulation will do.

Since there are no external relay contacts provided with the IC-502, I used a simple hand switch arrangement to key the amplifier. I connected a microswitch to a ten-foot cord which runs to the amplifier. When I am using the amplifier, I hold the push-to-talk mike in one hand, and the microswitch in the other. Alternatively, a foot switch could be used for this function. During standby periods, the bottom of the bias pot is lifted above ground by opening a pair of contacts on relay RLY1. This biases the grid of the 6146 into the cutoff region, which results in cooler operation of the final tube. Relay RLY1 is a twelve-volt DPDT type,

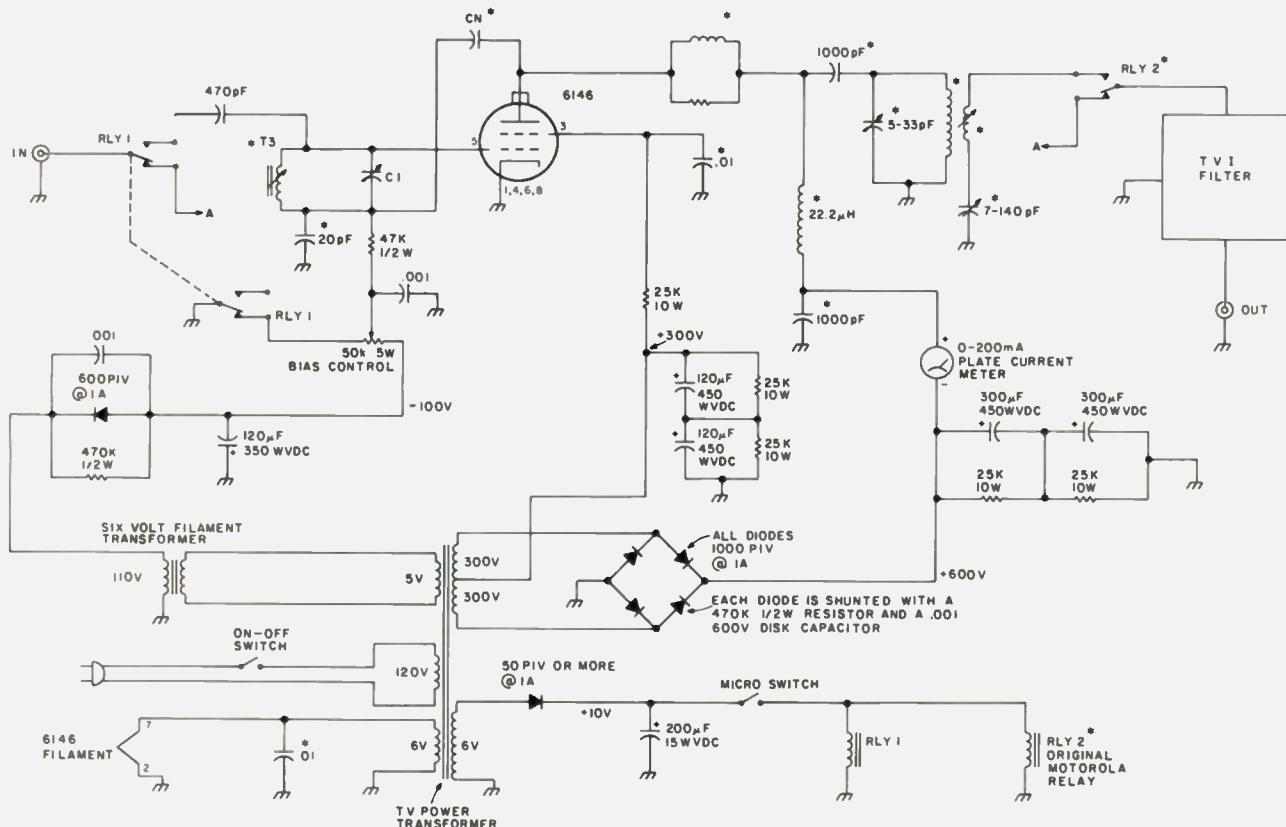


Fig. 1. C1 is an E.F. Johnson-type 148-4, 50-3.2-pF air-variable capacitor or equivalent. Note: All parts marked with an asterisk are from the Motorola transmitter strip which is a TA234 series. The relays are shown in the receive position. L—25-30 MHz; ML—30-35 MHz; MM—35-40 MHz; HL—40-45 MHz; HM—45-50 MHz; HH—50-54 MHz.

and RLY2 is the original antenna changeover relay in the transmitter strip. All of the rf wiring should be done with RG-58/U coaxial cable. Except for insulating the high-voltage filter capacitors, the rest of the wiring is noncritical.

Some sort of 6 meter TVI filter should be placed on the output side of the amplifier to ensure that it meets FCC harmonic suppression requirements.

Tune-Up

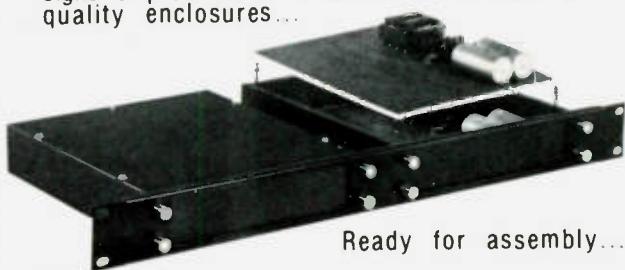
Without any rf drive applied to the amplifier, adjust the grid bias control so that the plate idle current is about twenty millamps. Next, connect an swr bridge between the amplifier and the exciter. Adjust the exciter to the frequency you most commonly use (such as 50.11 MHz), and place it in the tune position. In the case of the

IC-502, place the transceiver in the CW position, and put the key down. Then with the amplifier off, adjust C1 for the lowest swr. With the amplifier peaked for 50.11 MHz, I am able to QSY up to 50.4 MHz without any substantial drop in power.

Over the past few months, the on-the-air reports which I have gotten while using the amplifier have been gratifying. Stations that I have contacted on a regular basis report about a 4 S-unit improvement when the amplifier is turned on. Often, over long tropospheric bending paths of 400 miles or more, the amplifier has made the difference between whether I was able to make a contact or not. If you construct one of these 6 meter amplifiers, I am sure you will find it a valuable asset to your station. ■

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Look over these great issues on the following pages...

What have you missed?

68

JUNE 68. Surplus Issue! transformer tricks; BC-1206 RX; APS-13 ATV TX; low-voltage dc supply; surplus scopes; FM rig commercial xtal types; Wilcox F-3 RX; restoring old equipment; 75A-1 RX mods; TRA 19 on 432; frequency counter uses; transceiver power supply; uses for cheap tape recorders; surplus conversion bibliography; RT-209 walkie on 2m; ARC-1 guard RX; RTTY TX TU.

JULY 68. Wooden tower construction; tilt-over towers; erecting a telephone pole; IC AF oscillator; "dB" explained; ham club tips (part 1).

SEPT 68. Mobile VHF; 432 FET preamps; converting TV tuners; xtal oscillation stability; parallel-tube design; moonbounce rhombic; 6m exciter (corrections Jan. 69); 6m transceiver (corrections Jan. 69); 2m DSB amp; ham club tips (part 3).

NOV 68. SSB xtal filters; solid-state troubleshooting; IC frequency counter (many errors & omissions); "CV" transformers; space communications odyssey; pulsar info; thin-wire antennas; 40m transistor CW TX/RX; BC 348 double converter; multifunction tester; copper wire specs; thermistor applications; hi-voltage transistor list; ham club tips (part 5).

Available Issues published from 1968 are listed at the end of this catalogue.

JUNE 62. Surplus Issue! DMQ-2 Beacon TX on 220; increasing ARC-2 transceiver selectivity; PE-97A power supply conversion; BC-348 band-spread; inductance tester; converting BC-230 TX; beginner's RX using BC-453; receiver motor tuning; transistor CW monitor; BC-442 antenna relay conversion; mobile loading coils; increasing Two-er selectivity; TV with the ART-26 TX; TRC-8 RX on 220; ARC-5 HF RX & TX; ARC-3 TX on 2m.

AUG 63. Battery-op 6m station; diode noise generator; video modulation; magic T-R switch; antenna gain; halo mods; CW break-in; VEE beam design; coax losses; rf wattmeter; TX tube guide; diode power supply; "Lunchbox" squelch; swr explanation; vertical antenna info; Info on Windom antenna.

OCT 63. WBFM transceiver ideas; HF propagation; cheap phone patch; emotetuned vga; construction hints; antenna coupler; 35 vertical; filament transformer construction; 2m nististor converter; Lafayette HE-35 mcds; buyer's guide to RX & TX; product detector; novel converter; compact mike amplifier.

Available Issues published from 1960-1963 are listed at the end of this catalogue.

64

FEB 64. 2m multichannel exciter; RX design ideas; magic t/r switch; loudspeaker enclosures; 40m 2 W TX; look at test equipment; radic grounds; 40m ZL special antenna; neutralization.

Available Issues published from 1964 are listed at the end of this catalogue.

69

JAN 69. Suppressor compressor; HW 12 on 160; beam tuning; ac voltage control; 2m transistor TX; LC power reducer; spectrum analysis info; 6m transistor RX; operating console; RTTY autostart; calculating oscillation stability; low-power 40 CW TX; sequential relay switching; slightless operator's bridge; ham club tips (part 7).

FEB 69. SSTV camera mod for fast scan; tri-band linear; selective AF filter; unijunction transistor info; Nikola Tesla biography; mobile installation hints; Extra class license study (part 1).

MAR 69. Surplus Issue: TCS TX mods; cheap compressor/amp; RXZ calculations; transistor keyer; better balanced modulator; transistor oscillators; using blowers; half-wave feedline info; surplus conversion bibliography; Extra class license study (part 2).

APR 69. 2-channel scope amp; RX preamp; Two-er PTT; variable dc load; swr bridge; 100 kHz marker gen.; some transistor specs; SB-610 monitorscope mods; portable 6m AM TX; 2m converter; Extra class license study (part 3).

MAY 69. 2m turnstile; 2m slot; RX attenuator; generator filter; short vee; quad tuning; using antennoscope; measuring antenna gain; phone patch reg; swr Indicator; 160m short verticals 15m antenna; HF propagation angles; FSK exciter; kw dummy load; hi-power linear; Extra class license study (part 4); all-band curtain array.

JUNE 69. Microwave power generation; 6m SSB TX; 432-er TX/RX; 6m converter; 2m 5/8 wave whip; UHF TV tuners ATV video modulator; UHF FET preamps; RTTY monitorscope; Extra class license study (part 5); building UHF cavities; mini-vee for 10-20m; VHF vfo.

JULY 69. AM modulator; SSTV signal generator; 6m kW linear; 432 kW amp; 432-32 TX/RX; 6m IC converter; radio-controlled models; RTTY IC TU; audio notch filter; VRC-19 conversion; tube substitution; 2m transistor exciter; Extra class license study (part 6); HF FET vfo.

AUG 69. FET regen for 3.5 MHz up; FM crystal switching; 5/8-wave vertical; Introduction to ICs; RTTY tone generator; good/bad transistor checker; 2m AM TX; measure transistor Ft; 160m propagation; triac applications; simple i-f sweep generator; transistor keyer; SB-100 on 6m; xtal frequency measurement; Extra class license study (part 7); FM deviation meter; QRP AM 6m TX; circular quads; FM noise figure; transistor parameter tracer.

SEPT 69. Tunnel diode theory; magic tee; soldering techniques; wave-travel theory;

cable shielding; transistor theory; AM noise limiter; AFSK generator; transistor amp debugging; measure meter resistance; diode-stack power supply; transistor testing; 2½ W 6m TX; HX-10 neutralizing; capacitor usage; radio propagation; AM mod percentage; Extra class license study (part 8); 3-4002 linear; ATV vidicon camera; 2 transistor testers; FET compressor; rf plate choke.

OCT 69. Super-gain 40m antenna; FET chirper; telephone info; scope calibrator; thyrector surge protector; slower tuning rates; identify calibrator harmonics. FM adapter for AM TX; CB sets on 6m; proportional control xtal oven; xtal filter installation; Q-multiplier; transceiver power supply; Extra class study (part 9).

NOV 69. NCX-3 on 6m; I-f notch filters; dial calibration; HW-32A external vfo; 6m converter; feedline info; rf Z-bridge; FM mobile hints; umbrella antenna 432-34 TX (part 1); power supply tricks with diodes; transistor keyer; transistor bias design; xtal VHF signal generation; electronic variac; SB-33 mods; Extra class study (part 10); SB-34 linear improvements.

DEC 69. Transistor-diode checker; dummy load/attenuator; tuned filter chokes; band-switching Swan 250 & TV-2; 88 MHz selector; math exercises; RTL xtal calibrator; transistor PA design; HV mobile p.s.; 1-10 GHz frequemeter; CB rig on 6m; Extra class license study (part 11); 1970 buyer's guide.

70

JAN 70. Transceiver accessory unit; bench power supply; SSTV color method; base-tuned center-loaded antenna; 6m bandpass filter; Extra class license study (part 12); rectifier diode usage; facsimile info.

FEB 70. 18-inch 15m dipole; 6m converter; high-density PC board; camper-mobile hints; 2m frequency synthesizer; encoding/decoding for repeaters; DX-35 mods; panoramic VHF RX; variable-Z HF mobile mount; Extra class license study (part 13); linear IC info; QRP 40m TX; IC Q-multiplier.

MAR 70. Gdo applications; charger for dry cells; FM frequency meter; PC board construction; ham FM standards; cheap rf wattmeter; multifreq FM oscillation; "i-f" system modules (part 1); Six-er mods; gdo dip lite; Motorola 41 V conversion; CW monitor; buying surplus logic; SSQ-23A sonobuoy conversion; GRC-9 RX/TX conversion; Extra class study (part 14); intro to VHF FM.

APR 70. Noise blower; 2m hot carrier diode converter; repeater controller; understanding COR repeater; 7/8-wave 2m antenna; Extra class study (part 15); inexpensive semiconductors; renovating surplus meters; linear amp bias regulator; hi-performance I-f amp & agc system; SSB bfo for shortwave radio; vacuum-tube load box; general FM dope & repeater guide; meggering your antenna.

MAY 70. comments on "FM docket" #18003; future of CW; FM-AM RX aligner; 5/8-wave verticals; using 2m intelligently; auto burglar alarms; power supplies from surplus components; "i-f" system modules (part 2); VHF FET preamps; educated "Idiot" lights; postage-stamp 6m TX; Extra class study (part 16); Bishop IFNL; low-band police monitor; mobile CW TX; Witchita autopatch.

JUNE 70. DDRR antenna; vfo circuit; remote swr indicator; Indoor HF vertical; two RX on one antenna; environment & coax loss; 2-ell trap verticals; buying surplus; two 40m QRP TX; 21dB 2m beam; Extra class study (part 17).

JULY 70. Improved Color Slow-Scan Television; How to Build a Keyer; 450-MHz Mighty Mite—one-transistor superregenerative receiver; Cheapie 6-Meter Half-Gallon—use 811-As and be heard; A High-Performance Power Supply—using an IC voltage regulator; Latham Island DXpedition; Db to Power; Protection for Grid-Dipper Coil's; Mobile CW Receiver; QSLing... Ham Radio's Own Con Game.

SEPT 70. Integrated Circuit CW ID Generator; The Indication Oscillator—another dipper circuit, 1-400 MHz; Tuning VHF Receivers—clever infinite attenuator and oscillator unit; Repeater Antenna Separation; Diode Stacks; Deluxe Receiver Gain Control—using one transistor and a zener; Reed Relays for Coaxial Switching; Beer-Can Two Meter Coaxial Antenna; Converting 24 V Relays to 115 V ac; versatile 2m MOSFET Converter—low noise, high gain, ultra stable.

NOV 70. Differential J-FET preamplifier; Remote Quad Tuning; Two-Watt Six Meter Transmitter—using the crystal-heterodyne vfo; Semiautomatic FM channel Scanning; Low-cost Automatic Keyer—an excellent "first project"; Ac Switching with Self-Powered ICs—clever zero voltage switch; Pioneer Radio on the Prairies—what it was like 45 years ago; SST-1 Solid-State Transceiver for 40 Meters; A Low-Cost RF Wattmeter; Calibrate That Calibrator.

DEC 70. Solid-state VHF exciter; delta frequency control for SSB; 2m transistor FM TX; HW-100 offset tuning; "little gate" dipper; 3-5002 HF linear; General class study (part 5); "transi-test" (no good errors!); transistor p.s. current limiter.

Available Issues published from 1970 are listed at the end of this catalogue.

71

JAN 71. Split phones for DXing; Heath Ten-er mods; CW duty cycle; repeater zero beater; HEP IC projects; 10-15-20 parabolic ideas; lightning protection; IC RX accessory; attic antennas; double-balanced mixers; permanent marker tool; ham license study questions.

FEB 71. metal locator; varactor theory; AFSK unit; SSTV patch box; ATV hints; RTTY tuning indicator; tone encoder/decoder; 220-MHz converter; SSTV magnetic deflection; IC code oscillator; 6m TX beeper; General class study (part 6); RTTY intro; perfboard terminal; low ohmmeter.

MAR 71. IC audio filter; IC 6m converter; trap vertical ideas; digital counter info; surplus equipment identification; HF linear; simple phone patch; repeater audio mixer; digital RTTY accessories; coax-hanger ground plane; General class study (part 7).

MAY 71. 75m mobile whip; 2m preamp; transistor amp design; 10m DSB TX; portable FM transceiver directory; audio compressor-clipper; transistor .M frequemeter; 450-MHz link TX; simple Af filter; 1-tube 2m transceiver; surplus 2m power amp; General class study (part 8).

JUNE 71. 2m beam experiments; 3-ell 2m quad; multi-band dipole patterns; weather balloon vertical; pocket-pager squelch; Two-er vfo; tuning mobile whips; transistor power supply; capacity decade box; 40m gain antennas; General class study (part 9).

JULY 71. IC audio processor; audio signal generator; CW filter; 2m FM oscillator; 2m collinear vertical; FM supplier directory; Motorola G-strip conversion; transistor beta tester; General class study (part 10).

AUG 71. Ham facsimile (part 1); 500-Watt linear; dimensions for July collinear; 4-tube 80/40 station; vfo digital readout; Jupiter on 15m; General class study (part 11); pink ticket wave-meter.

SEPT 71. Transformerless power supplies; solid-state TV camera; IC substitution; two rf wattmeters; IC compressor-agc; multichannel HT-200; ham facsimile (part 2); causes of man-made noise; vfo with tracking mixer; General class study (part 12); transistor heatsinking; IC pulse generator; phone patch isolation; hcd wattmeters.

OCT 71. Emergency repeater COR; transceiver power supply; predicting meteor showers; digital switching; reverse-current battery charger; passive repeaters; earth grounds; audio "tailoring" filters; Swan 350 mods.

NOV 71. 3-ell 75m beam; motor-tuned ground plane; 2m gain vertical; transistor biasing; split-site repeater; fox-hunting; audio filter; transistor/diode tester; xtal tester; 6m kW amp; 10-15-20m quad; transistor pi-net final; antenna feedline; communications dBs; 2300-MHz exciter.

DEC 71. Convert Your 7-MHz Cubical Quad to All Bands; The Indoor Quad; Get-

65

Available Issues published from 1965 are listed at the end of this catalogue.

66

Available Issues published from 1966 are listed at the end of this catalogue.

67

MAY 67. Quad Issue! 432 quad-quad-quad; expanded HF quad; two el quac; mini-quad; 40m quad; quad experiments; half-quad; three-el quad; quad bibliography; FET vfo; tube troubleshooting; HF dummy load; understanding "dB"; HF SSBI/CW RX; geometric circuit design; GSB-201 transceive; FET converter for 10-20m; hi-pass RX filters.

JULY 67. VE ham radio; VE0 hams; DSB adaptor; home-brew tower; transistor design; '39 World's Fair; ground plane antenna; G4ZU beam; SSTV monitor; UHF FET preamps; IC "I-f" strip; vertical antenna; V-F/UHF dipper; tower hints; scope monitoring; operating desk; S-line cross-band; hi-school ham club; Heath HR-10 mods.

OCT 67. HF solid-state RX; rugged rotator; designing slug-tuned coils; FET converter; SSTV fix generator; VHF log-pe lodes; rotatable dipole; gamma-match cap; old-time DXing; modern DXing.

Available Issues published from 1967 are listed at the end of this catalogue.

ting to Know Tes Squared Ell; More Power From 6146s; Radio Direction/Range Finder; Morse Memory—30-letter memory for ident, contests, etc.; SCR Mobile Theft Alarm, DX QSOs or contacts; Code Short-hand; VHF Double Sideband.

72

FEB 72. A Solid-State High Frequency Regenerative Receiver—e cheapo using one IC; Tips for Raising Your Code Speed to 20 WPM; Why Not Try QRP? VHF Dummy Load Wattmeter; CW DX On ½ Watt—enjoy QRP with this 1-Watt rig; 20-60 W 1-4 Band TX—two-tube CW transmitter; Quick and Easy PNP/NPN Transistor Sorter; Self-Contained Reflected Power and CW Monitor; Circuits, Circuits, Circuits; The Automatic Transmission Line Tuner.

MAY 72. Quick Band Change Mobile Antenna—with output indicator; How to Get the Stuff into the House; Anti-CW RTTY Autostart; A Modern VHF Frequency Counter—can be built for under \$100; TV Sync Generator—using ICs; Radio Astronomy; Noise and Receiving Antennas; The Sewerpipe Antenna—2m FM, of course; Circuits, Circuits, Circuits; Simple Car Ammeter—all solid state.

JUNE 72. Six Elements on Twenty Meters—eliminate QRM; Slow Scan Television—basics; Active Filter Design and Use—all kinds of filters ... Part I; Radio Astronomy for Amateurs (Part II); 20 dB Beams—design and construction of VHF antennas; Phasing Multiband Antennas—ten thru eighty meters; 300-MHz Frequency Scaler—extends frequency counters to VHF; Circuits Circuits, Circuits; RTTY Filters—elliptic function filters; Troubleshooting for the Novice.

JULY 72. Solid-State VHF Amplifier; The Phase-Locked Loop; VHF Converters; Add \$15 T-Power; 1296-MHz Mixer; The VHF Specialists FM Amplifier; Meteor Shower DXing; Tone Decoder and Carrier Relay Circuits—using the 741 op amp; Flying Spot Scanner for SSTV—solid-state unit, simple, relatively; Active Filter Design, Part II.

AUG 72. SSTV info; speech processor; FM repeater info; test probe construction; GE Proline ac supply; 432 rf testing; preamp compressor; Six'er mods; phone patch; Two-er info; solar info; SCR regulator for HVPS; "ideal" xtal oscillator; FM RX adapter; auto theft alarm.

SEPT 72. Plumbicon TV camera; WWVB 60-kHz RX; cigar tube signal generator; CW active filter; rf testing at 1296-3500 GHz; balun antenna feed; transistor power supply; IC 6m RX; IC FM/AM detector (part 2); active filter design (part 3); K2OAW frequency counter (part 3); 2m frequency synthesizer (part 1).

OCT 72. Corrections for Aug. FM RX adapter; 2m frequency synthesizer (part 2); 6m transistor vfo; nano-ampere meter; time-frequency measurement (part 1); active filter design (part 4); repeater timer; Extra class Q&A (part 3); balloon vertical; D generator; time-delay relay; 432 filter deas; dc-ac inverter; hc-diode converter; RTL decade and nixie driver; plus-minus supply for ICs.

NOV 72. RF transistor power amps; RTTY Selcal; IC trr RX transistor keyer; emergency power; 220-MHz preamp; double-delta antenna; simple converter using modules; HF rf tester; "lumped line" oscillator; 2m frequency synthesizer (part 3); K2OAW counter errata; 2m preamp; Extra class Q&A (part 4); hi-Z voltmeter; Nikola Tesla story; VHF swr meter; transistor regen RX; 432 SSB transverter; ac arc welder; intro to computers; hybrid AM modulator; HR-10 RX mods; 10m transistor AM TX; 40m ground plane; IC logic demonstrator; overload protection; i-f/rf sweep generator; digital frequency counter; aural TX tuning.

DEC 72. SSTV scope analyzer; 2m FM RX; tone burst encoder and decoder; universal i-f amp; autopatch hookup; LM380N info; voltage variable cap info; 2m 18-Watt amp; SSB modulation monitor; xtal freq/activity meter; 10 A var. dc supply; transmission line uses; radio astronomy; inductance meter; 75 to 20m transverter; LED Info; 40m preamp; transistor vfo; 1972 index; 2m preamp.

73

JAN 73. HT-220 touchtone; 3-el 20m yagi; 50-MHz frequency counter; speech processor; 2-tone generator; FM test set; tilt-over tower; 2m converter using modules; tunable AF filter; six-band linear; 10m i-f tuner; diode noise limiter; CWSSB agc; HW-22A transceiver 40m mod; HAL ID-mod.

MAR 73. A Fast-Scan Facsimile System—use it with SSTV; Six and Two Meter High Power—using a \$25 surplus amplifier; A Digital Tape Distributor for RTTY; The Amplifier—all band, 1200 Watts; Popular SSTV Circuits (Part II); Improving the Indoor Antenna System—using copper foil; FM Deviation Meters, Time Frequency Measuring System (Part III); Another Use for 400-Cycle Transformers; Bandpass Filter Design.

APR 73. FM deviation meter; 2m FET preamp; two 2m power amps; repeater control (part 1); repeater licensing; European 2m FM; FM scanner adapter; RCA CMU15 mods; lightning detector; CB alignment; gate; transistor rf power amps (part 2); repeater economics.

JUNE 73. 220-MHz signal generator; UHF power meter; repeater licensing info; RTTY autoswitch; 40m hybrid vfo TX; antenna polar mount; 10-15-20m quad; K2OAW counter mods; double coax antenna; ham summer job; tone decoder; field strength meter; nicad battery pack ohmmeter; FCC regs (part 1).

JULY 73. Tuneable Oscillators for 2m FM Receivers; Basic ATV System—a T-44 transmitter strip does most of the work; Multiple Output Frequency Standard—lets you calibrate your receiver in .0625-Hz increments; Digital Identification Unit 450-MHz Power Divider—easily constructed matching system for stacked arrays; CW Filters, Bared and Compared—complete with scope traces and bandwidth specs; 85 dB Gain 2m Antenna Compromise Multiband Antennas; Grid-Dip Tuning the Quad Antenna.

AUG 73. Log-periodics (part 1); tone burst generator; rf power amp design; transistor radio intercom; 160m antenna; SSTV monitor; low-cost frequency counter; VOM design; QRP 40m TX; 432-MHz exciter; FM audio processing; FCC regs (part 3).

SEPT 73. Repeater control system; log periodics (part 2); 2m RX calibrator; PLL IC applications; TT pad hookup; Heath HW-7 "S" meter; OSCAR-6 Doppler; 2m coaxial antenna; 2m converter; IC keyer; measure antenna Z; FCC regs (part 4).

NOV 73. 450-MHz exciter; intro to ATV circuits; nicad voltage monitor; autopatch connections; IC meter amplifier; TR-22 ac supply; indoor vertical; IC AF filter; momentary power failure protection; 160m antenna coupler; Motorola HT info; SSTV-IB, Class B AF amp; FCC regs (part 6).

74

FEB 74. SSTV monitor info! IC audio amps; scope sweep generator; 15/20m vertical; telephone line control system; PC board construction; var-Q AF filter; blown-fuse indicator; 40m CW station with Ten-Tec modules; simple preamp module; HF rf tester; "lumped line" oscillator; 2m frequency synthesizer (part 3); K2OAW counter errata; 2m preamp; Extra class Q&A (part 4); hi-Z voltmeter; Nikola Tesla story; VHF swr meter; transistor regen RX; 432 SSB transverter; ac arc welder; intro to computers; hybrid AM modulator; HR-10 RX mods; 10m transistor AM TX; 40m ground plane; IC logic demonstrator; overload protection; i-f/rf sweep generator; digital frequency counter; aural TX tuning.

APR 74. VOX for repeaters; tone-operated relay; HF transverter; 10 to 2m TX converter; remote control panel for scanner; RCA FM TX tuning; subaudible tone generator; FCC regs (part 9); repeater atlas.

MAY 74. CD car ignition; audio compressor info; interference suppression for boats; auto burglar alarms; 2m IC preamp; 10m FET converter.

JUNE 74. Poor Man's Quad; Reconciling the Long Squared Quad—developing a new type antenna; Antenna Load Indicator; Matching; Remotely tunable Antenna Coupler; A Practical Ground System for 160; Wide-Range Antenna Tuner; Old Antennas and New Baluns—build a double zepplin; A Multiband Ground Plane—10-40 meters; Mod Quad for Frustrated Cliff Dwellers.

JULY 74. 4-1000A linear; universal frequency generator; universal AFSK generator; 555 IC timer; 80m phased array; 135-kHz/432-MHz preamps; 10m QRP AM TX; 3000 V dc supply; how to read diagrams.

AUG 74. Toroidal directional wattmeters; 450-MHz FET preamp; use gdo to find "C"; Trilineline TT pad hookup; R390 & R392 RX mods; tracking CW filter; aural voltmeter; universal regulated supply; SSTV scan converter; TTL logic problems; ID timer.

SEPT 74. MOSKEY electronic keyer (part 1); WX warning system; Heath IO-103 scope mods; QRP 6m AM TX; rf speech clipper; audio noise limiter; WX satellite on SSTV monitor; universal IC tester; miniature rig construction; tower construction; infinite rf attenuator; electronic photo flash ideas; IC "select-oject."

OCT 74. Microtransistor circuits; synthesized HT-220 (part 1); repeater government; regulated 5 V dc supply; FM Selcal; removable mobile antennas; Motorola metering; 2m vertical collinear; Motorola model code; 2m coaxial dipole; 1.6-MHz i-f strip; MOSKEY electronic keyer (part 2); carbon mike circuit; hi-power lo-pass filter; 6m preamp; 3-wire dipole; ATV sync generator; NCX-5 mods; mobile whip for apartment dwellers; SSTV automatic vertical trigger.

NOV 74. K2OAW counter update; regulated 5 V dc supply; wind direction indicator; synthesized HT-220 (part 2); 20m 3-el beam; autopatch pad hookups; double-stub antenna match; Novice class instruction; digital swr meter (part 1); 6m converter (1.6-MHz i-f); "C-bridge"; MOSKEY electronic keyer (part 3); Aug. SSTV scan converter errata; repeater off-frequency indicator.

DEC 74. Care of nicads; wind speed/direction indicator; WX satellite video converter; electronic keyer; hints for Novices; unknown meter scales; SSTV tape ideas; TTL logic probe; public service band converter; tuned-dipole test receivers; digital swr meter (part 2); telephone pole beam support; rhombic antennas. 1974 Index.

75

FEB 75. Heath HO-10 scope mod for SSTV; electronic keyer; digital satellite orbital timer; OSCAR-7 operation; satellite orbital prediction; Heath SB-102 mods; comparing FM & AM; repeater engineering; Robot 80-A SSTV camera mod; neutralizing Heath SB-110A; "Bounceless" IC switch; tape keyer for CW TX.

MAY 75. IC Callsign Generator; Playing with Power on 432; Does Ether Cause Gravity? OSCAR Your FM Rig; In Pursuit of the Perfect SSTV Picture; Ac Power for the HW-202; You Can Work 75m DX; The Postage Stamp Squeicher; disaster in Honduras.

JUNE 75. Home Brew this SSTV Monitor; El Cheapo Superbeam; The Smart Alarm; RF Power at 432; Dirt Cheap Tunable I-F for Converters; All Band Frequency Marker; Front Burner for Six; Three on Fifteen; Presto! Transistor Checker from VOM; How to Put on a Professional Slide Show.

JULY 75. OSCAR Special! Antennas for OSCAR—What Really Works?; How You Can Take OSCAR's Temperature; FM Alignment Oscillator; The Audio Synthesizer for RTTY, SSTV and Whatever; Ham Radio in the Arctic—1925; Gee, What's a Zepp?; Vertical Antennas for the Novice; Preventing Regulator Carnage; The Ultimate in Variable Selectivity; Phone Patching—A Public Service.

AUG 75. 146/432-MHz helical antennas (part 2); 20 minute ID timer; digital swr computer (part 1); debugging rf feedback; DVM buyer's guide; WX satellite monitor; CMOS "accu-keyer"; PC board methods; sweep-tube final precautions; compact multiband dipoles; small digital clock; accessory vfo for HF transceiver; modern non-Morse codes; multi-function generator; 2m scanning synthesizer errata; KP-202 walky charger; 10m multi-element beam.

SEPT 75. Calculating frequency counter; WX satellite FAX system (part 1); IC millivoltmeter; three-button TT decoder; troubleshooting SSTV pix; 40m DX antennas; 146/432-MHz helical antennas (con-

clusion); digital swr computer (conclusion); red relay for CW bk-lin; NE555 preset timer; power-failure alarm; portable QRP rig power unit; precision 10 V dc reference standard; 135-kHz i-f strip; telephone handsets with FM strip; Motorola T-44 TX mod for ATV; 0-60-MHz synthesizer (part 1); ham radio PR.

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76

FEB 76. Build a Starfleet Communicator—Trekies special; Synthesized IC Frequency Standard! You Can Make Photo PC Boards; How's Your Speech Quality? ASCII-to-Baudot converter; RTTY Autocall—the Digital Way; Improving the FT-101; Night DXing on 10 and 15m; Really Soup Up Your 2m Receiver; Put Your SB-10 on 160m.

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78

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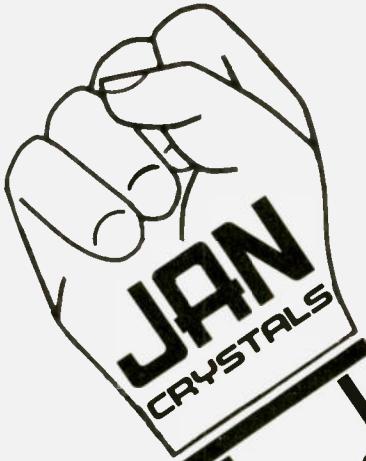
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The Big Bopper

As an interested amateur, you probably have heard many on-the-air and off-the-air conversations discussing phased antennas. If you have never operated a phased array, you may have wondered—with honest skepticism—whether they really perform well. In particular, are

they generally equivalent to other good arrays and do they have any unique advantages? The answer to both questions is yes. Obviously, there are some qualifications. No single approach to an antenna system can ever be the universal solution to every ham's radiation problem.

But if what you need is an effective radiation system on multiple bands with a low physical profile and you want it to be relatively inexpensive as well as easy to construct and maintain, then you may want to consider a vertical array.

The two-tower vertical

array described here has directional radiation (and reception) characteristics which are easily rotatable. The system is sturdy, attractive, and readily usable on five bands, with no outdoor adjustments required while operating. Standing-wave ratios are

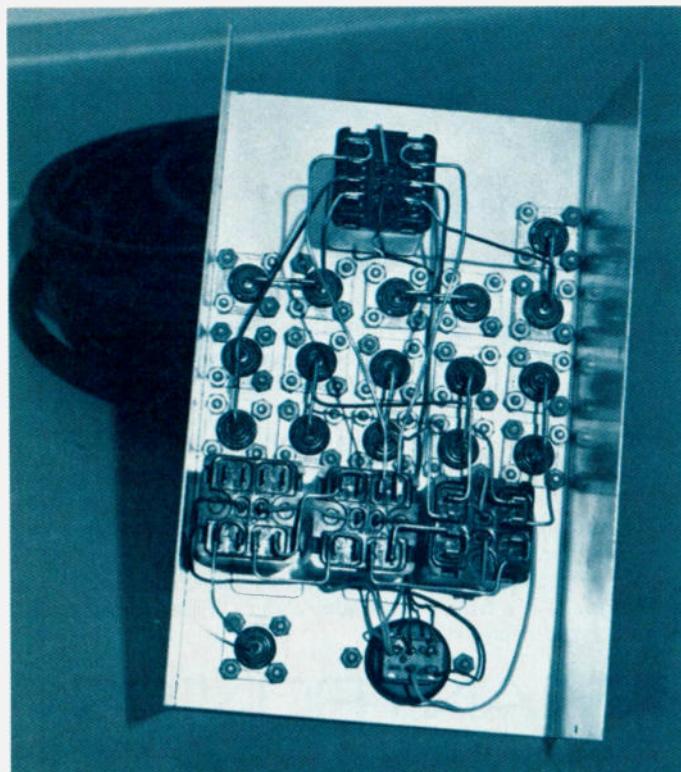


Photo A. Wiring side of the relay switch box.

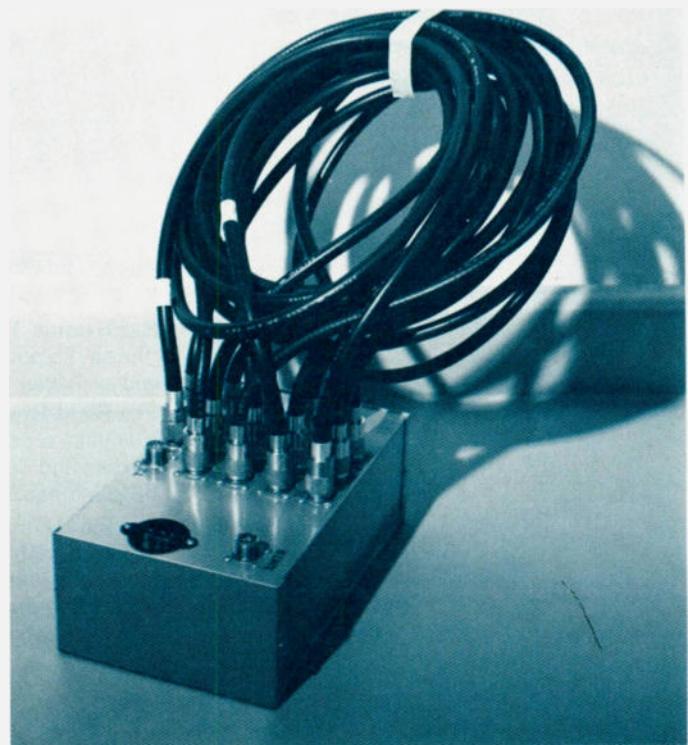


Photo B. Relay switch box "buttoned up" with seven "internal" lines attached.

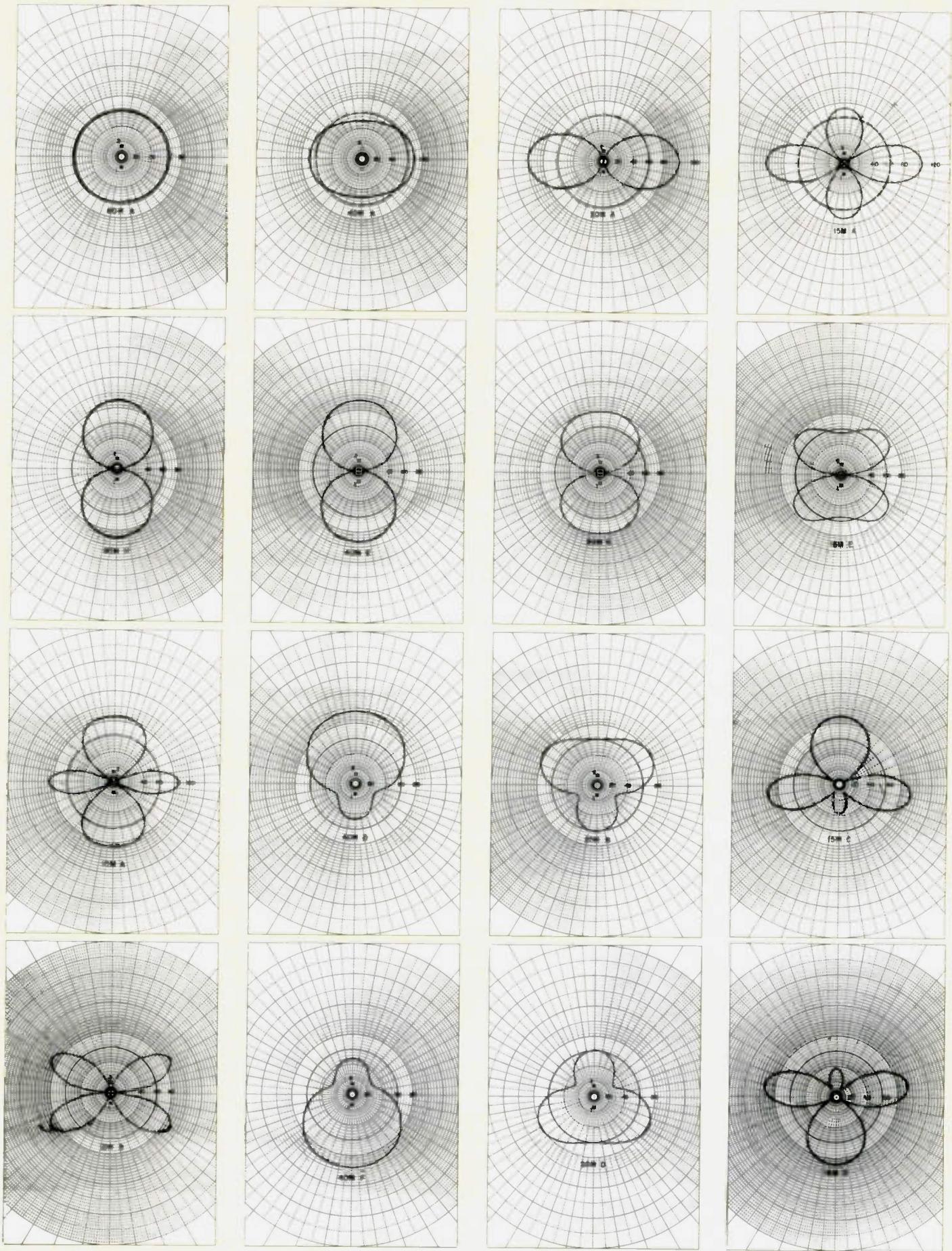


Fig. 1. Idealized radiation patterns. All patterns represent field intensity from 100 radiated Watts. Dots 1 and 2 represent towers 1 and 2. Capital letters represent switch positions. Circles = 61.8 millivolts/meter at one mile. This is the field intensity from a single tower with the same 100 Watts radiated.

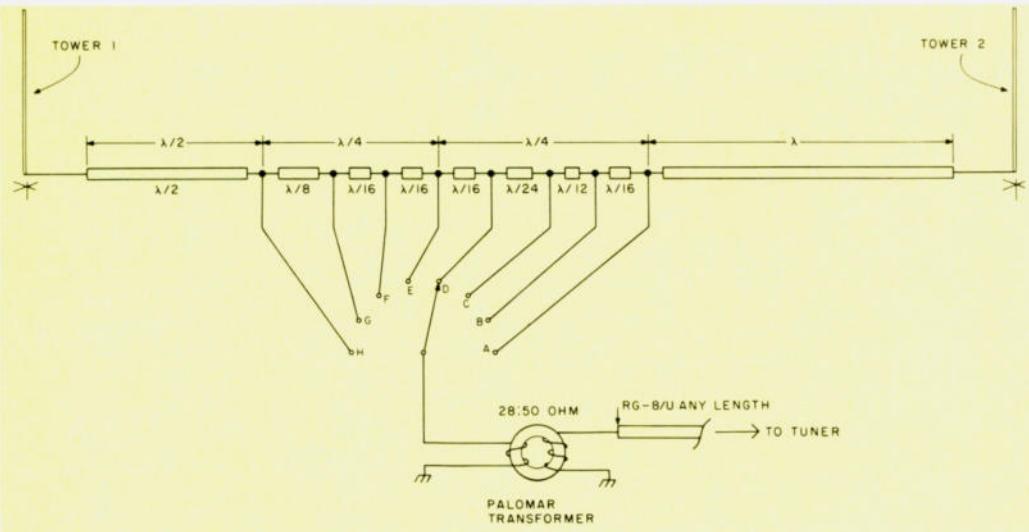


Fig. 2. Block diagram. All notation refers to 40 meters only. All cable and delay lines are RG-8/U or a 50-Ohm equivalent. The transformer is a Palomar Engineers or equivalent (see text). Towers and trapped units on each band are Hustler 4-BTV or equivalent.

moderate under all conditions of operation. The transmission line can be any convenient length. An antenna tuner or matching network is recommended between the transmitter and the antenna system.

The effects of mutual impedance on the array operation have been determined. The radiation patterns have been calculated for typical amateur-level

field intensities. Importantly, the patterns recognize the effects of any unbalanced tower currents which sometimes result. The construction of the array, and its operation, is completely straightforward and uncomplicated. It does not require great resources, either real estate or budget. It may be especially attractive if you have been denied a more

conventional beam because of zoning restrictions, house construction limitations, or general appearance considerations.

Since the tower separation is fixed, the array cannot be equally directional on all five bands. Offsetting this quality is the ability of the system to operate on five bands with the turn of an indoor switch. It is emphasized that this abili-

ty to operate on all major bands with variable directional enhancement on each band is a capability that is not a commonplace feature of typical antenna systems. Also, as will be seen, once the system is tuned to a pattern, it is very tolerant of frequency shifts within a band and can be characterized as having quite a flat load response.

There is no special design requirement for the individual trapped towers. For symmetry, it is recommended that they be nearly the same. Commercial units are readily available and are competitively priced. If your machine shop facilities are limited, buying the verticals may be the quickest and least expensive approach. On the other hand, if you are more venturesome, or able, various means of trapping are described in the amateur handbooks and periodicals which most of us have in our ham libraries. Actually, you may have one vertical antenna already. In this situation, the array discussed may offer you an easy extension in operating capability, with only incremental cost or effort.

This array effectively uses all of the basic lobe patterns that are intrinsic to a two-tower system with a common feedline. Intentionally omitted are some familiar two-tower patterns—perfect cardioids, for instance—that can be generated only with an elaborate feed system so as to maintain balanced tower currents.

The Design Approach

Vertical antennas—towers—used as the basis of the design approach have a number of appealing qualities. They are unobtrusive in residential neighborhoods. Vertical towers with tuned traps allow easy band changing and multiband operation.



Photo C. The desk control unit in place.

AT-120 and AT-180

Antenna Tuners recommended for TS-120S and TS-180S All Solid-State HF Transceivers.

Why is the use of an antenna tuner so much more important with the TS-120S and TS-180S all solid-state HF rigs than with transceivers having tube-type final amplifiers?

Tube-type final amplifiers generally handle a broader range of load impedance than a transistor final. However, RF power into the antenna system will decrease with an increasing impedance mismatch, and tube life may even be shortened if the mismatch is extreme. Transistor final amplifiers, on the other hand, require a 50-ohm nonreactive load for efficient power transfer and are not very tolerant of high SWR. Therefore, protection circuits are used in the TS-120S and TS-180S to reduce RF power output significantly under high-SWR conditions, thus preventing damage to the solid-state devices.

In the TS-120S, an SWR detection circuit detects reflected-wave voltage, which is then amplified and applied to the ALC circuit as a protective voltage to control power output. Thus, as SWR increases, RF power output decreases continuously.

In the TS-180S, the final amplifier functions normally up to an SWR of 3:1, at which point the protection circuit drops RF power output significantly.

Using an antenna tuner such as the AT-120 to match the TS-120S or the AT-180 to match the TS-180S will lower the reflected power at the transceiver to avoid detection by the protection circuit, thus enabling the transceivers to produce full RF power output and even with rigs which have tube final amplifiers, we recommend an antenna tuner (such as the AT-200 to match the TS-520 or TS-820 Series) for optimum coupling to antennas with high SWRs.

A major advantage of using an all solid-state rig such as the TS-120S or TS-180S is the elimination of final-amplifier tuning and loading. It's great to be able to switch bands, dial up any frequency, and transmit immediately, especially when operating mobile or in a contest or chasing DX. Isn't this advantage lost if an antenna tuner has to be used?

We recommend using an antenna that has a low SWR (below 1.5:1) and that presents a proper impedance match (50-ohms) to the transceiver. Then the full advantages of using an all solid-state rig can be realized. Furthermore, the antenna will be more efficient, and power will not be reflected back to an antenna tuner.

However, many antennas are not broad enough to cover an entire band, and may have an SWR below 1.5:1 in just a portion of the band. The antenna may be cut for resonance in the middle of the portion of the band that is mostly used. When operating outside this portion, where SWR exceeds 1.5:1, the antenna tuner should be switched in.



AT-120

AT-180

Therefore, with a well-designed antenna, the antenna tuner may be switched out for most operating, and the full advantage of using a no-tune all solid-state rig may be realized. But for those occasions when operating in the band portions where the antenna is not resonant and reactance increases or when, for some reason, the antenna develops a high SWR or a poorly matched antenna is used, the antenna tuner should be switched in to obtain full RF power output.

During those occasions when the antenna tuner is needed with an all solid-state rig, it would probably be advantageous with a tube-type rig also for optimum power transfer to the antenna system. With a tube-type final, plate tuning and loading adjustments would be required in addition to adjusting the antenna tuner. With an all solid-state transceiver, only the antenna tuner would need adjusting during those occasions when it is required.

What are the primary features of the AT-120 and AT-180 antenna tuners?

The AT-120 antenna tuner is very compact (only 6 inches wide, 2-3/8 inches high and 6-1/4 inches deep) - perfect for mobile mounting with the mounting bracket provided - and operates on 80 through 10 meters. It consists of an antenna coupler and an SWR meter (which can be illuminated). Although much smaller, it complements the appearance of the TS-120S.

The bandswitch has a "THROUGH" position for switching the AT-120 out of the circuit. Input impedance (to the transceiver) is 50-ohms and output impedance (to the antenna system) covers 20 to 300 ohms, unbalanced. It handles 150 watts (120 watts on 80 meters). The SWR meter measures from 1.0:1 to 10.0:1.

The AT-180 antenna tuner matches the TS-180S (same height) and consists of a through-line watt and SWR meter, antenna selector switch, and, of course, an antenna coupler.

It operates on 160 through 10 meters, with a 50-ohm input impedance and an output impedance of 10 to 500 ohms (10 to 400 ohms on 160 meters), unbalanced. Switches allow up to 20 or 200 watts of forward or reflected power to be measured. (It is not intended for use at the output of a linear amplifier.) UHF-type connectors are provided for the input, two antenna outputs, and a dummy load, and a standoff connector is provided for a wire feedline.

With both tuners, the "R TUNE" (for resistance component) and "X TUNE" (for reactance component) controls are adjusted alternately with a CW carrier applied until minimum SWR or reflected power is obtained.

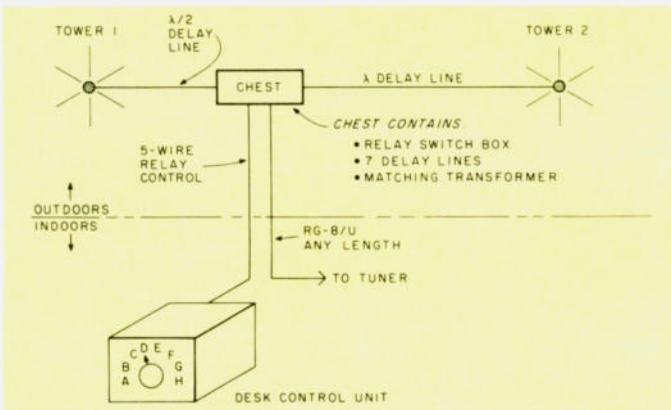


Fig. 3. Bird's-eye view showing the simplicity of the system.

The low-angle radiation patterns from verticals enhance DX probabilities for a given transmitter power. The absence of high-angle vertical lobes usually equates to effective use of radiated power. Properly constructed, ground-mounted verticals can have low losses and are structurally sound in almost all weather.

For 40 meter and 80 meter operation, a two-tower vertical system is one of the few practical options that can be used to enhance directivity. This is especially true if rotation of the pattern is desired. Typically, at these frequencies, the beam element lengths in conventional arrays become too ponderous for most hams to cope with.

Finally, in the two-tower system, no mechanism or external rotators are required. The quick switching of the lobe patterns is especially handy for reception when operating in a net or group. Often the compass bearing between received stations is quite wide. Quick optimization of the received signal is a real operating advantage. Sometimes, too, it is helpful to be able to partially null an interfering signal.

How This Antenna System Operates

The two towers are physically separated by $\lambda/4$ on 40 meters (34'5"). This is

a separation distance that will comfortably fit on many home lots. There is an eight-position switch at the operator's position that selects between the lobe patterns (Fig. 1). For example, switch position D on the 40 meter band selects a paddle-shaped pattern directed off the top end of the axis. Switch position F on the 40 meter band is an identical paddle-shaped pattern, but directed off the bottom end of the axis. The absolute geographic direction is determined by the tower axis orientation.

It is worth noting again that the division of current between the towers affects the pattern generated. Yet the current division between towers, with a common feedline, is a function of the tower self-impedances—including mutual impedance—as trans-

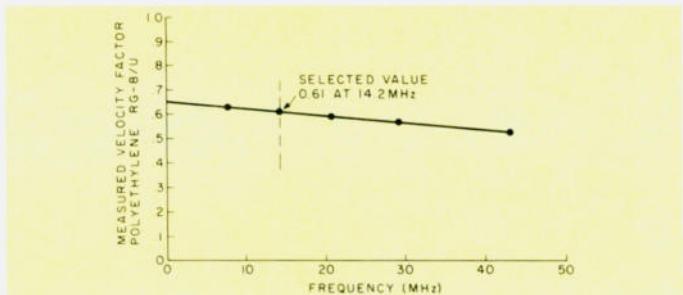


Fig. 4. Velocity factor versus frequency for polyethylene RG-8/U.

formed through the delay lines to the feed-in point. The individual tower impedances, in turn, for a given current phasing, are a function of the current division between them.¹ Sound circular? For example, it is possible with some patterns and phasings to initiate a multivibrator-type action. As a tower current goes down, the tower impedance goes up, which causes the current to go down further . . . etc. The resulting radiation pattern degenerates toward a single tower pattern or a circle. It can be speculated that such action has discouraged unsuspecting experimenters in the past when actual performance compared poorly with anticipated pattern enhancement.

These degenerative conditions have been avoided in this array. But to avoid them, some two-tower patterns, as shown in a *Radio Engineering Handbook* or

similar source, are not available in practice. Uncritical referencing to tables of these patterns is a common error. Such handbook displays are normally predicated on towers which are $\lambda/4$ high and which have equal currents in each tower. Maintaining equal currents under all conditions of operation and switching is not an easy trick. Instead, the lobe patterns in Fig. 1 are based on the unequal currents which occur normally. The patterns give good emphasis and direction selection on each band. By orientating the tower axis with care, good world coverage can be obtained.

All patterns as shown in Fig. 1 are "idealized," meaning that they are computed, not measured, and are assumed to be generated over perfectly conducting earth. For most ham installation planning jobs, these assumptions are quite acceptable. The

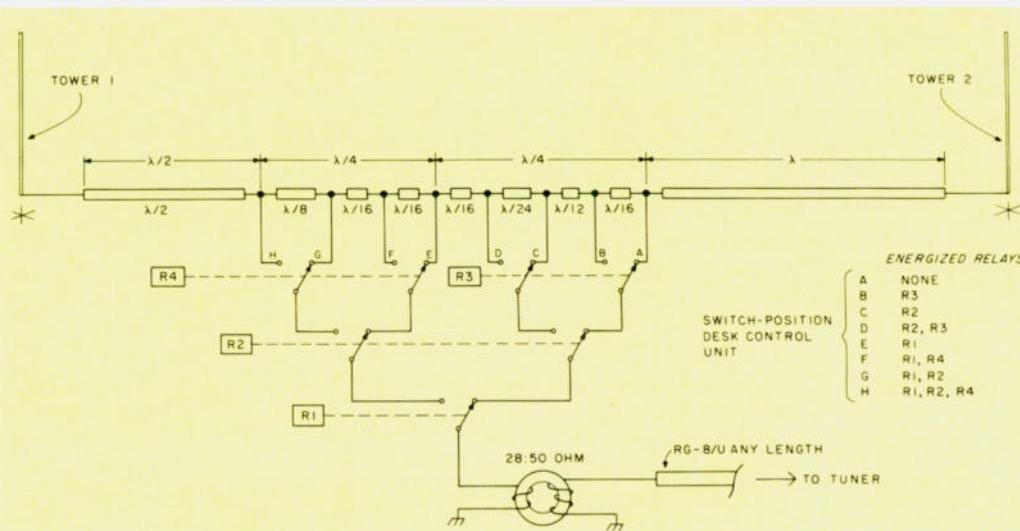


Fig. 5. Relay circuit. All relays are shown at rest. All wavelengths refer to 40 meters only.

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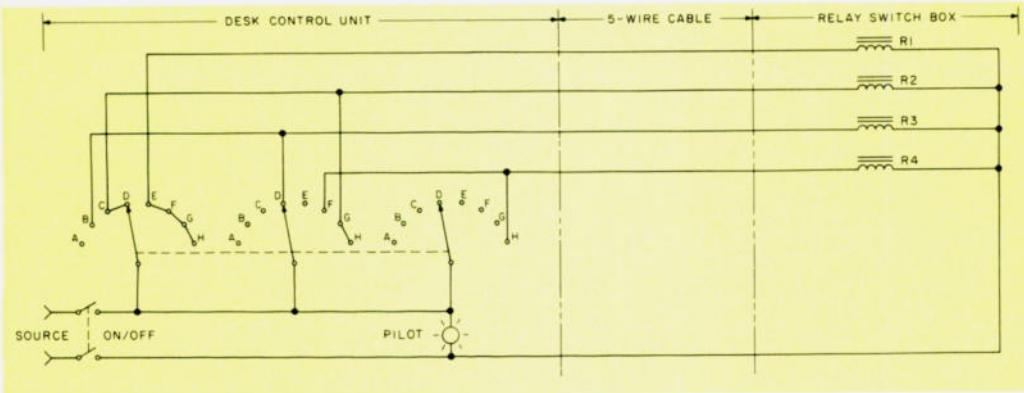


Fig. 6. Relay coil-control circuit.

charts are calibrated in millivolts/meter for a radius of one mile with a total rf radiated power of 100 Watts. To estimate the field strength for a different radiated power, P_2 , it is only necessary to multiply the chart reading in the chosen direction by $\sqrt{P_2/P_1}$. In our case, P_1 is 100 Watts. The circles in the charts represent 61.8 millivolts/meter at one mile, which is the field intensity that theoretically would be obtained if the 100 reference Watts were to be radiated from a single vertical tower.

The Block Diagram

The antenna system is shown in block diagram form in Fig. 2. For consistency, all notations in Fig. 2 and throughout this article relating to wavelength or delay degrees will refer to 40 meter operation. Conversion to other bands is then straightforward.

Although the towers are physically $\lambda/4$ apart (90°), they are separated elec-

trically by two wavelengths of delay line in nine unequal sections. The electrical centerpoint is at A. The feed-in point of the transmission line is determined by an eight-position switch.

In the D position, as shown, there is a $13/16\lambda$ delay to tower 1 and a $19/16\lambda$ delay to tower 2. In other words, the current in tower 2 lags the current in tower 1 by the difference in delays, or: $19/16\lambda - 13/16\lambda = 6/16\lambda = 3/8\lambda = 135^\circ$. At the same switch position, consider 20 meter operation. All electrical dimensions are now doubled. The towers are physically separated by 180 electrical degrees. Tower 2 current now lags tower 1 current by twice the amount at 40 meters, or 270° .

The eight-position switch function could be done manually as is illustrated. Such manual switching would be quite awkward in most operating situations, since the switch would necessarily be outdoors. Instead, the equivalent function is done remotely with four relays. The relay circuit and the matching transformer will be discussed in a later section.

The specification of trapped antennas for the towers is important. For predictable radiation pattern generation, it is desirable that each tower be an electrical quarter wavelength high for each

operating band (at least no higher than $5/8 \lambda$). Trapping achieves this condition. It may be tempting to try to use untrapped 40 or 80 meter quarter-wavelength towers for all bands. Such untrapped tower heights, when used on high frequency bands such as 10 or 15 meters, can cause confusion and difficulty. In effect, each tower can become a resonant "long-wire" with the associated intricate radiation pattern. Lossy high-angle lobes can result. The interaction of these complex patterns in a switchable two-tower array is apt to result in unpredictable and disappointing results without careful analysis.

Construction Features

One real advantage of this array is its easy and straightforward adaptation to construction. For instance, the full λ length of delay line to tower 2 and the $\lambda/2$ length of delay line to tower 1 permit the physical grouping of a switch box plus seven of the delay lines conveniently away from both towers. This statement allows for the physical shortening of the delay lines to accommodate transmission line velocity factors. This arrangement flexibility is often useful for appearance or screening considerations (Fig. 3).

While on the subject of velocity factors, it may be helpful to review my own

experience. Perhaps I should have known better, but I was unprepared for calibrated dip-meter measurement results, which indicate that the velocity factor in polyethylene RG-8/U varies with frequency (Fig. 4). A review of the technical literature that I have, plus manufacturer's data on hand, discloses no mention of this quality. A single standard "nominal" value for the velocity factor is usually given. Other experimenters may also have this gap in their information. Obviously, knowledge of this variable is significant if the same cable is to be used for generating specific delays, but at different frequencies—as in this array. I selected a velocity factor of 0.61 obtained at 14.2 MHz. This is a compromise value. Errors introduced on other bands are within reasonable limits with this value. Table 1 lists the physical lengths I have used.

It is suggested that the λ length of delay line, and also the $\lambda/2$ length of delay line, be cut a little long. Then, with all of the delay lines attached to the switch box but not connected at the towers, balance and trim the lines from midpoint A outward with a dip meter. This balancing will minimize any discontinuity effects in the switched leg due to relay points or coax hardware.

The Relay Switching Circuit

The functional switch in Fig. 2 is replaced by the relay circuit in Fig. 5. The switching relays are conventional medium-power units and need not be the expensive coax type. They should have two sets of SPDT contact points. A compact physical design is suggested. It is important that they have sufficiently heavy-duty contact points

Electrical wavelength	Physical length (vel. fac. = 0.61)
λ	84.8"
$\lambda/2$	42.4"
$\lambda/8$	10.7"
$\lambda/12$	7.5"
$\lambda/16$	5.35"
$\lambda/24$	3.625"

Table 1. Physical lengths of delay line sections. *Initially cut long and trimmed to final length with a dip meter (see text).

to be able to handle the currents involved. A contact current rating of 10 Amps or better will be adequate in most applications. Obviously, the relays should have good insulation and a low inductance switch path. Contact resistance should not be a problem since full rf current flows through any contacts that are used.

The relay coil-control cable is a five-wire cable. Since the cable runs from the shack to the switch box outside, it is likely to either be buried or at least to touch the earth. As a result, conventional safety precautions require that the relay coils should either be of the low-voltage type or else should be isolated with a suitable transformer arrangement.

In my installation, the relays are housed in an aluminum box, $3\frac{1}{2} \times 6 \times 10$ inches (Photos A and B). This box also serves as a mount for the SO-239 panel receptacles which make the connections to the transmission line and delay lines. This arrangement affords good shielding, short connections, and easy assembly and disassembly. Another plus is the ability to make easy substitutions of delay line lengths for experimentation. The layout is not critical and your own design should work well if attention is given to short lead lengths.

Selection of the relays is controlled from a desk control unit which basically consists of a 3-pole, 8-position rotary switch housed in a decorator metal box (Photo C). A pilot light and an on/off switch can be convenient additions. The relay coil-control circuit is detailed in Fig. 6.

Radials and the Ground System

All too frequently, hams underestimate the need for

a proper ground system with vertical towers and attempt to skimp by. This is false economy and wastes effort and power. An adequate radial system is a fundamental and necessary component of the total radiation array and should not be neglected. A working rule of thumb is to have a minimum of six radials per band with a minimum length per band of 0.125λ (Photo D).

The Transmission Line, Delay Lines, and Matching Considerations

Although simple in physical appearance, operation, and construction, there are some deceptively involved aspects to the electrical analysis of the array:

- There are two towers with mutual coupling which affects the individual tower impedance.
- Pattern switching affects the load impedance at each tower.
- The towers are not always operated at resonance with the corresponding introduction of reactance of varying magnitude and sign.
- The impedance of the system, viewed from the radial switch, is a resultant of two parallel paths—one to each tower. The length of these paths, their individual impedances, and the current division between them change with switching.

It should be clear that the system—considered as a load—varies under normal operating conditions. For the most part, these operating conditions, while

Band	Tower 1	Tower 2
10	1.30	1.25
15	1.45	1.50
20	1.25	1.15
40	1.20	1.20
80	1.00	1.05

Table 2. Swrs measured at tower base and at tower resonant frequency.

tedious to analyze, are quite solvable with the charts and equations of G. H. Brown, the liberal use of Smith charts, plus the determined application of patience and time with a programmable hand calculator. Fortunately, in practice, the system analysis complexity can largely be ignored while achieving excellent operating results. Anyway, the analysis is mostly needed for determining the pattern shapes when the tower currents are unbalanced. Still, it is realized that most users will want an insight into the operation without the tedium. The following discussion is intended for that purpose.

The range of impedances

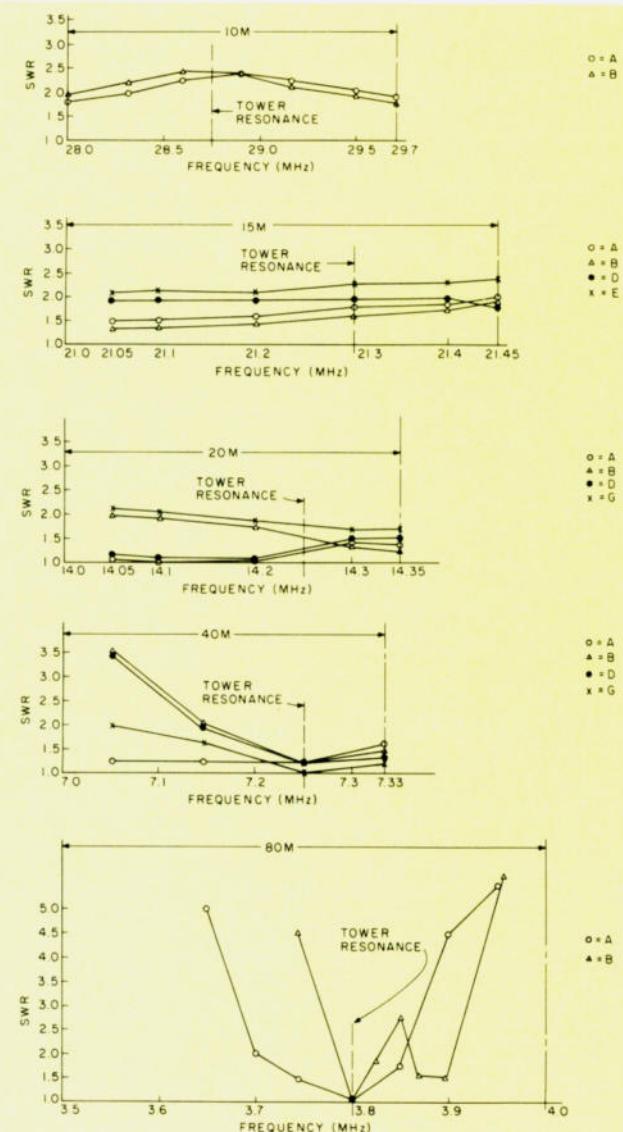


Fig. 7. Swrs versus frequency.

of the delay line. From switch position A, with 20 meter operation, there are two wavelengths of delay line to each tower. Since a half wavelength of transmission line repeats its input at its output, the tower 1 resistance is repeated at point H and again at point E. Similarly, tower 2 resistance is repeated at A. The electrical distance from E to D is $1/8\lambda$. The electrical distance from A to D is $3/8\lambda$. The question now becomes what happens at D? Whipping out our handy Smith chart, we find that the resistance of tower 1 will be seen at D as 49 Ohms resistance plus 12.5 Ohms of inductive reactance. Likewise, the resistance of tower 2 will be seen at D as 49 Ohms of resistance plus 12.5 Ohms of capacitive reactance. Since the reactances are equal and opposite, they will cancel. (This action is true at any tap. Its value will be discussed later.) As in example 1, this leaves us with two resistive legs in parallel. The net parallel resistance is 24.5 Ohms. This resistance, seen

half wavelength of line repeats its input at its output, the tower 1 resistance is repeated at point H and again at point E. Similarly, tower 2 resistance is repeated at A. The electrical distance from E to D is $1/8\lambda$. The electrical distance from A to D is $3/8\lambda$. The question now becomes what happens at D? Whipping out our handy Smith chart, we find that the resistance of tower 1 will be seen at D as 49 Ohms resistance plus 12.5 Ohms of inductive reactance. Likewise, the resistance of tower 2 will be seen at D as 49 Ohms of resistance plus 12.5 Ohms of capacitive reactance. Since the reactances are equal and opposite, they will cancel. (This action is true at any tap. Its value will be discussed later.) As in example 1, this leaves us with two resistive legs in parallel. The net parallel resistance is 24.5 Ohms. This resistance, seen

through the transformer, will appear as 44.1 Ohms, which, in turn, will generate an swr of only 1.13.

Unfortunately, when mutual impedance is reintroduced into our assumptions, the reactive components do not always cancel at all times. Also, the self-resistance of the towers may become different, which would mean that the currents will no longer be the same in each tower. Nevertheless, the function of the transformer can be understood as an impedance leveler and, in fact, practical experimentation is easy.

The transformer itself is a ferrite core unit. I use the Palomar transformer at the nominal ratio of 28:50 Ohms. Roll-your-own transformer winders should have little difficulty achieving the 1.0:1.8 ratio or close to it.^{2,3} The turns ratio is not excessively critical, and some experimentation with ratios may be rewarding. It has been my experience that swrs for a favored band can be improved, but usually at some expense to another band.

Adjustments/Tuning

The first operation in adjustment is to tune the towers individually to the same frequency on each band. Other things being equal, a midband resonant frequency will probably give better overall results than a frequency near band edge. Assuming the absence of an rf bridge, probably the most direct alignment technique is to measure the swr at the base of each tower and adjust for minimum. To avoid mutual coupling effects, the other tower should be removed from its standard and placed out of the immediate area.

It will be remembered that an ideal vertical $\lambda/4$ tower has a resonant resis-

tance of about 36 Ohms. Thus, with a 50-Ohm transmission line, the swr for an ideal tower would be $50/36 = 1.4$. Some hams are confused about this point—probably by advertising claims—and expect an swr of 1.0 or close to it.

The swrs that I measure on the transmission-line side of the matching transformer are shown in Fig. 7. These moderate swrs represent no threat to RG-8/U, even at full legal power. Coax losses at these ratios are insignificant. Most antenna tuners can easily cope with these ratios and will present a perfect load to the transceiver or linear amplifier.

Table 2 lists the swrs I measure at the base of each tower. Superficially, it would appear that I have the best adjustment on 80 meters where the swrs are the lowest. The contrary is the case. My towers are telling me that I have an inadequate radial system on 80 meters. In effect, I have a dissipative ground resistance in series with the radiation resistance, and, therefore, the total resistance is raised. The net resistance at 80 meters nearly matches the transmission line characteristic impedance. Although this raised overall resistance gives a nice low swr, it also indicates that I am wasting some valuable rf power heating the ground!

Perhaps of more interest is the shape of the swr curves in Fig. 7. For the most part, 80 meters excepted, they do not have the familiar lopsided, saucer-shaped, resonant curves. Instead, they tend to be gently curved and be level, or have a slight tilt. That is, the swr does not change appreciably with change in operating frequency! This is a feature of the system. Stated another way, once an antenna tuner is adjusted for a par-



Photo D. The radial system. The aluminum box is handy, but not necessary. It is used to weather protect a PL-239, which was used for cable length experimentation.

ticular lobe pattern, at a particular frequency, it is usually unnecessary to re-adjust it much while operating across a significant band segment. For emphasis, it should be noted that the chart for the 10 meter band has a 5:1 compression in its frequency scale when compared to the other bands. Its flatness is thus even more impressive.

For an explanation of these unusual SWR curve

shapes, recall that, in our second example, there was a cancelling of reactance at each switch tap, providing that there is an integral number of half wavelengths of delay line between towers. If the operating frequency is moved away from the resonant frequency of the towers, an identical reactive component will be introduced at each tower. These reactive components will tend to cancel at

the taps, leaving only resistance. It is like having a built-in compensating tuner!

In addition to matching the antenna system to the output pi network of your transceiver or linear amplifier, an antenna tuner gives added protection against spurious radiation. This antenna system is an excellent radiator on all bands at all times! It will not filter nor inhibit any undesired harmonics. An

antenna tuner is an investment in tranquillity. ■

References

1. G. H. Brown, "Directional Antennas," *Proceedings of the Institute of Radio Engineers*, January, 1937. See equations 37, 38, 39, and 40 on page 92.
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3. Jerry Sevick, "Broadband Matching Transformers Can Handle Many Kilowatts," *Electronics*, November 25, 1976.

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V70	144-148MHz	10-15W	75-90W	216x330x178mm	11.7 kg (26 lbs)	No	\$315.00
V71	144-148MHz	1.3W	75-90W	216x330x178mm	11.7 kg (26 lbs)	No	\$349.00
V180	144-148MHz	5.15W	170-200W	216x330x178mm	13.5 kg (30 lbs)	CW & FM	\$539.00
V350	144-148MHz	10.20W	350-400W	432x330x178mm	23.4 kg (52 lbs)	Yes	\$895.00
V130B	220-225MHz	10-15W	70-85W	216x330x178mm	11.7 kg (26 lbs)	No	\$329.00
V135B	220-225MHz	25-35W	140-160W	216x330x178mm	11.7 kg (26 lbs)	CW & FM	\$469.00
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F220		Fan Kit, 230VAC		135x135x50mm	1 kg (2.2 lbs)		\$ 33.00
*F135		Fan Kit, 115VAC		381x140x89mm	3.2 kg (7 lbs)		\$ 59.00
*F235		Fan Kit, 230VAC		381x140x89mm	3.2 kg (7 lbs)		\$ 59.00
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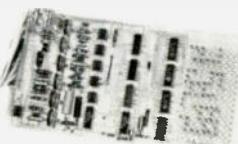
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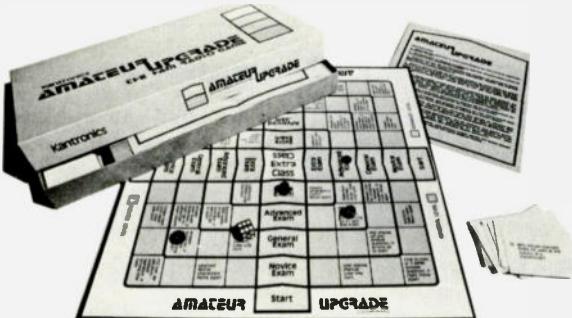
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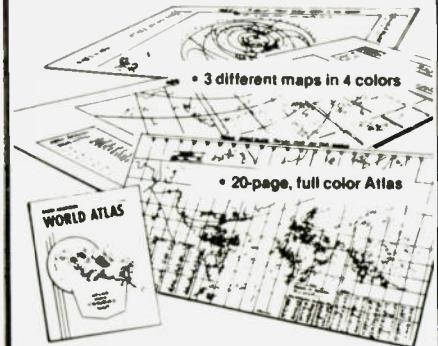
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Contests

from page 30

RST) and 3-digit serial number starting with 001.

SCORING:

European stations count 1 point per QSO on any band; non-European stations score 1 point per QSO on 20 through 10 meters, and 3 points per 80/40 meter QSO. Count every call area in the above-mentioned countries on every band as a multiplier, e.g., LA1 = LB1 = LJ1 and SM3 = SK3 = SL3. A portable station in Norway or Denmark counts as the 10th call area, e.g., W2XXX/OZ counts for OZ0. Some countries have no geographical call areas, but count this contest as if they had. The sum of all complete QSO points from all bands multiplied by the sum of multipliers from all bands is the final score.

AWARDS:

Certificates awarded to the highest-scoring station in each participating country and US call area in each operating class separately on CW and phone. A special SAC-plaque will be given to the continental winners on both CW and phone. A reasonable score is requested for both certificates and plaques. Depending on the number of contestants in each country, the contest committee will consider more certificates.

ENTRIES:

Write separate logs for CW and phone. The logs are to be filled in in the following order: date and time in GMT, station worked, sent and received message, multiplier, and points. Separate logs for each band are recommended. On the summary sheet the contestant will write his/her callsign, name and address, the final result, the operating class, and a signature that he/she fully agrees to the rules. The logs must be mailed no later than October 15 and sent to: SRAL, SAC Contest Committee, PO Box 306, SF-

00101 Helsinki 10, Finland. The decisions of the contest committee are final and definite and they reserve the right to change the rules. Usual disqualification rules.

DELTA QSO PARTY

**Starts: 1800 GMT September 29
Ends: 2400 GMT September 30**

All amateurs are invited to participate in the 10th annual party which is sponsored by the Delta Division of the ARRL. Amateurs outside of the Delta Division (Arkansas, Louisiana, Mississippi, Tennessee) will attempt to contact as many stations inside of the Delta Division. Delta Division stations will attempt to contact as many stations both inside and outside of the division. Stations may be worked on each band mode; portables and mobiles may be reworked on the same band mode if they change counties.

FREQUENCIES:

CW – 3550, 7050, 14050, 21050, 28050.
SSB – 3990, 7290, 14290, 21390, 28590.

Novice – 3725, 7125, 21125, 28125.

EXCHANGE:

QSO number, RST, and QTH (ARRL section for non-Delta Division; county and state for Delta Division).

SCORING:

Delta Division stations score number of QSOs times the number of ARRL sections (75 maximum); outside Delta Division score number of QSOs times the number of counties worked (316 maximum). DX stations may be worked by Delta Division stations but do not count as multipliers.

ENTRIES AND AWARDS:

Logs must include date/time, station worked, exchange, band, emission, and multiplier. Logs must be postmarked no later than October 21 to be eligible for award consideration. Logs will be returned if request-

ed. Send logs to Malcolm P. Keown W5XX, 213 Moonmist, Vicksburg MS 39180. All amateurs contacting 5 different stations in each of the 4 states comprising the Delta Division will receive a certificate. Section awards to the 3 highest-scoring stations in each Delta Division state, 4th and 5th place if warranted; section awards to high-scoring stations in each ARRL section and country outside Delta Division, 2nd and 3rd if warranted. Plaques to high-scoring stations both inside and outside of the division and to high-scoring portable and mobile stations in the division. Another plaque to the high-scoring Delta Division club station (no limit to number of operators or transmitters, but all QSOs must be made from same QTH).

COLLEGE RADIO SCRIMMAGE

**Starts: 1900 GMT September 30
Ends: 0100 GMT September 31**

Entry classes: 1) alumni and 2) club station. Open to all amateurs; alumni may work other alumni and club stations. Entry may have one or more operators but must have only one transmitter (no distinction between single- and multi-operator).

EXCHANGE:

Name of college, junior college, or university you last attended and the last two digits of the year you graduated, will graduate, or last attended. Club stations substitute the words "Amateur Radio Club" for number. Non-collegians substitute the words "High School" for college name, but are not eligible for awards and may not be counted as a multiplier.

SCORING:

Stations may be worked once per band. Multiply total QSOs by number of different colleges worked (college counts once regardless of band).

FREQUENCIES:

1815, 3895, 7230, 14280, 21355, 28560.

AWARDS AND ENTRIES:

Trophy for top scorer in each entry class. Certificates as well. Logs must be received by November 1 to be eligible. Send an

SASE for results to: Penn State Amateur Radio Club, K3CR, 202 Engr. Unit E, University Park PA 16802.

FALL CLASSIC RADIO EXCHANGE

**Starts: 2000 GMT Sunday, September 30
Ends: 0300 GMT Monday, October 1**

This contest is sponsored by the Southeast Amateur Radio Club, K8EMY, of Cleveland, Ohio, and is open to all. The object is to restore, operate, and enjoy older equipment with like-minded hams. A classic radio is any equipment built since 1945 but at least ten years old (an advantage, but not required in the Exchange). The same station may be worked with different equipment combinations and on each mode on each band. General call is "CQ CX" and non-contestants may be worked for credit.

EXCHANGE:

Your name, RST, state/province/country, receiver and transmitter type (home brew send PA tube type, e.g., "6L6").

FREQUENCIES:

CW – up 60 kHz from low band edges.

Phone – 3910, 7280, 14280, 21380, 28580.

Novice/Tech – 3720, 7120, 21120, 28120.

SCORING:

Multiply total QSOs by number of different receivers, transmitters, states, provinces, and DX countries worked on each band and mode. Multiply that total by your classic multiplier: total years old of all receivers and transmitters used, three QSOs minimum per unit. If equipment is a transceiver, multiply age by two.

ENTRIES AND AWARDS:

Certificates and other appropriate memorabilia are awarded periodically for highest scores, longest DX, best excuse, and other "unusual achievements." Send logs, comments, anecdotes, etc., to Stu Stephens K8SJ, 2386 Queenston Rd., Cleveland Heights OH 44118. Include an SASE for a copy of the Classic Radio Newsletter.

Ham Help

I would like to correspond with any ham using a Heath HW-18-3 160-meter single sideband transceiver on CW as well as SSB.

William Toben W7JGL
1244 W. Schafer Drive
Tucson AZ 85705

I have been taking 73 Magazine now for three years. I want my Novice license very much. I am 70 years old and

blind – my wife reads the magazine to me. I have 73's code tape and know the code very well. However, I can't receive too well, as I don't know Braille and can't read back anything after writing it down.

I believe I could pass my theory test. I need to know just what is required for the blind to pass the test and if there is anything else I could get that might help me. I have radios

that I listen to all the time, but I would be happy to talk to someone.

Can anyone give me any help or advice?

Raymond L. White
3341 James Avenue
Fort Worth TX 76110

I am looking for meter rewinding information and/or specifications for a replacement meter for a Johnson Viking Challenger transmitter. Thank you.

George L. Smyth N8AMZ
1316 Eastern Ave.
Morgantown WV 26505

I need any information I can get for a Keithley Milliohmometer Model 502.

Will buy, copy, or pay to have copied information on batteries, operation, calibration, and schematics.

Dave Hammer K0PCL
610 E. South
King City MO 64463

I would like to form an exchange of solar, alternative energy, and anti-nuclear information with thinking hams.

James G. Coote WB6AAM
906 Dexter Street
Los Angeles CA 90042

W2NSD/1 NEVER SAY DIE

editorial by Wayne Green

from page 6

... not just for our own egos, but in order to keep newcomers entering our hobby and thus keeping it alive.

Without the support of a national organization, what can we as individuals or even as clubs do to help the situation? Well, getting publicity is not difficult once you recognize how the system works ... ask your terrorist or other politically active group. No, I'm not suggesting that hams start fire bombing police cars; I'm just pointing out that a realization of how television and newspapers select what they think is news will give us some ideas which can allow us to take advantage of a seriously unbalanced situation.

Take a large parade ... maybe thousands of people and floats. That's not news. But if you add six yelling and screaming agitators in a small group jostling the parade, that's what you'll see on the news that evening ... and not the ten thousand people in the parade. No, we don't want to break up parades either ... we just want to understand what *is* news and what *is not* news.

Our getting publicity should take advantage of amateur radio's possibilities. For instance, suppose we set up a national net to check food prices and report the inequities. Then amateurs could set up cars or vans in front of grocery stores with a big report poster showing the comparison between the prices in that store and other stores around the country ... where large price differences are found.

There are plenty of these price inequities to be found, and the project would make television all over the country. Heck, right here in Peterborough you can find 10% to 20% differences in the prices of identical products just between the Grand Union and A&P stores! As long as stores end prices in a 9, we know that these prices are not honest ones ... at least the chances are 90% against their being honest. We know for darned sure that a price ending in 9 is 90% of the time picked because it seems lower than one which is one cent more. If products were honestly priced, the prices would end in a random selec-

tion of numbers ... which should reflect the cost of manufacture and distribution. We also know darned well that if the honest price of a product comes out to be, say, 80¢, it will turn up on the store shelf at 89¢, not at 79¢.

There are fewer and fewer products which are not being priced in 10¢ steps, all ending in a 9. The carpet people go in dollar steps, always managing to end in 99. Bless 'em.

Ham communications could do the job, using two meters for price reports in the immediate area and perhaps 20m for long-range comparisons. And if you don't think something like that wouldn't make every television news broadcast in the country, you have another think coming.

Hams working to beat down the artificially high prices on food ... hams working to bring honesty to pricing and show up the price gougers.

The gas shortage may not last very long, but we could take advantage of the tremendous interest in this situation by setting up a ham reporting network on two meters in each area which could let motorists know where gas was to be found, where the gas lines were the shortest, and gas prices ... particularly on weekends. Temporary signboards along the major routes, manned by hams (with coordination by a local ham club), would make the news broadcasts. The gas stations could be checked by telephone for the latest news of their situation or checked out by mobile units. It would be a fun project and would provide a real service. Now that the idea is out, if you don't do it, the CBers will, and you'll have to see them on television and in the newspapers. Your choice.

Those are just a couple of ideas. I'm sure you'll be able to come up with more and possibly better ones. Those came

to me while I was taking a shower this morning. Take a shower. Let's get amateur radio known and appreciated.

THE ARRL LINE

Every time I tune from 14,200 on up past 14,275 to where the channels are packed solid with nets and QRM, I tend to think of the dividing line at 14,275 kHz as being the ARRL line. This dividing line is there because the League forced the FCC to take away General frequencies. Before the advent of "incentive licensing," there was no such line and General Class licensees had the use of the entire band. Now they're jammed up into half the phone band, away from 99% of the attractive DX ... certainly away from the relatively open frequencies in the lower part of the band ... and from SSTV experimenting.

This all started in 1963 when the League, being run at the time by some old-timers who wanted the ham bands to be back the way they were before the War, pushed through a disastrous piece of legislation. Amateur reaction to the proposals was so furious that it almost stopped the League, despite their inside help via Herb Hoover, Jr. The amateur reaction brought the sale of amateur equipment almost to a halt and stopped the growth of the hobby for over ten years. Talk about disaster!

This was the time when most of the famous names in ham manufacturing disappeared ... Hallicrafters, National, Hammarlund, Johnson, Harvey Wells, B&W, etc. Tuning past 14,275 brings back these memories of the havoc wrought by the League in the 1960s. '73 was too small then to be more than a cry in the wind, but I think that without the cry the League proposal to remove Generals entirely from the major DX bands would have gone through. The retaining of half of the bands by Generals was better than nothing.

The result of all this? Proof that forcing amateurs to upgrade would not work. Many of us held that the FCC should offer rewards such as better call letters rather than taking away frequencies and then giving them back if you upgraded. If you are locked above 14,275, it's something to think about.

SAROC ZOMBIE

That's what they call the dead brought back to life ... or at least some semblance of life ... and it seems apropos for the SAROC hamfest. This is a strictly commercial venture run for profit by a chap in Nevada and it died a couple years ago of extreme neglect. I see a report that this exploitation of amateur



CQ . . . DE LEAVENWORTH

The chap in the middle, between the two police officers, has some serious problems. The photo was taken from a video monitor shortly after the arrest of the unhappy chap in the center for shoplifting.

This fellow had previously been into the store, one of the largest ham stores in the country, and had been suspected of gaining a good deal of weight during his earlier visit. This time they were waiting for him, complete with video monitoring and recording. The police were called and arrived to watch him stuffing his coat with books and parts via the TV monitor.

You see him in the Polaroid above, just after having been read his rights and as he was being carted off to the police station for booking.

The monitoring system has worked well, providing the staff with some rather clear pictures of shoplifters for use during later visits, complete with police in the audience. The tape recorder allows a moment by moment replay of the surveillance at any later time . . . which helps in the case of any later claims of innocence.

radio has been announced again for next year in Las Vegas.

The main purpose of SAROC is to provide a legal excuse for a few of the ham manufacturers to go to Las Vegas for a vacation in January. The actual business done is a farce. For instance, one major supplier brought in a truckload of ham gear for his exhibit at the last show and over the three days of the show sold \$1,500 worth of gear. That's less than one tenth of the minimum acceptable business for him just to break even.

Now, if the show was that lousy when there was still plenty of gas for the California hams to drive over the desert to come, who will be there when gas is in short supply and a lot more expensive? Ten years ago, when FM and repeaters were firing everyone's imagination, the local Vegas FM club ran a hospitality suite. Art Householder of Spectronics also had a hospitality suite where he made everyone crazy with desire for recycled Motorola HT-220s which he pulled from his pockets and sold in a kind of under-the-table atmosphere. It was heady stuff and FMers were king of the roost.

No more. The last show, two years ago, was held together mainly by itinerant hams in town for the concurrent winter Consumer Electronics Show and who came over to the hamfest for a few minutes to see what their poor relatives were doing. Next year, CES will be January 5-8th. The hamfest will be the 10-12th and the Fifth Annual Ham Industry meeting will be in Aspen, Colorado, on January 12-19th. I'll probably hop out on the 7th for a couple of quick days at CES, visit the Wilson factory on the 9th in Vegas, see who was wealthy enough in the industry to "exhibit" at the hamfest, and then poop or up to Aspen for the yearly industry conference and workshops.

SAROC, being run strictly for a profit, has a tremendous charge for attending . . . I think I paid \$17 last time . . . and for that I got to wander around a handful of booths. It took me maybe half an hour to see everything. There was no recognizable program. ARMA held a meeting, but many of the manufacturers were unable to find the meeting room, which was very well hidden.

Now, why is Wayne Green so down on SAROC? Is it that I'm getting even because I haven't been asked to speak? No . . . as a matter of fact, I've been asked to speak and have refused. It's a long story and there is a lot of background. It comes down to a question of supporting what I

feel are unsavory elements. I sure wish someone would take the time to write an exposé of the people behind this hamfest.

Anyway, between the gas problem and the CES show ending two days before the hamfest, it seems unlikely that the next SAROC will be any improvement over the last one . . . which had CES to help and no gas problems . . . and still was a disaster. Perhaps this commercial show should be put to rest and remembered for its heyday ten years ago instead of its more recent disappointments.

DXERS AND LISTS

A lot of true-blue DXers are very uptight over list operations. For those of you who have not enjoyed the spice of working serious DX, a list of calls is taken by some intermediary and passed along to the rare DX station, who then works the list and only the list. This can be very upsetting for some operator who comes along after the list has been drawn and finds that while he can hear the DX station easily, he can't work him. Oh, bitter anguish!

The operators of rare DX stations have several choices open to them when they get on the air. First, they know that a normal contact is out of the question because every time they stand by there will be dozens, then hundreds, of stations all calling them simultaneously. The din soon gets so great that it is impossible to hear any call distinctly and the tendency then is to wait out the caterwauling until there is just one station left to come back to him.

The encouraging of "tail-ending" quickly screws up everything. The waiting hordes quickly detect what is happening and there soon develops a two or three operator contest to see who can be the last one to call. None of them can even hear the DX station and none will quit the bedlam, so the calling can go on for several minutes. One W5 I know got steamed up at a certain K6 and the two of them refused to budge or listen on the frequency for well over an hour. The poor DX station had gone to bed long before they stopped jamming each other.

Some DX ops like the system where a DX station operates from just outside the US band and tunes for calls. I like this, too, for it doesn't take long to find out where the chap is tuning and jump in the pileup on that frequency.

Having been on the other end of the pileups, I have some observations from the viewpoint of the rare DX station. Newcomers, I've DXed from the following rare and relatively rare spots: JY, OD, 5Z, YK, YA, VU, 9N, FK, VR2, 5W, KS6, FO8,

KW6, KG6, KC4, HH, PJ, 4U, OH0, VP9, VP7, and probably some more that don't come right to mind. The main problem in working a lot of stations from a rare location is not in hearing them — it's in pulling the calls apart and getting the individual calls.

After trying every known system for working 'em as fast as possible, which included working by call districts on my own frequency, working from outside the US band and listening on one frequency in it, tuning a small band of frequencies, and working from lists, there was no question . . . the list could beat everything by far.

Well, let me amend that. The list situation works great if you manage to get hold of some sharp operators to gather the lists for you. I somehow always managed to get one or maybe two good ops and one thoroughly-going idiot. I really hate it when I think I have everything set up for some really fast operating. I've rounded up three chaps with good signals to move off my frequency and copy down call letters for me . . . maybe ten at a time. Then all they have to do is come on my channel and read off the list quickly so I can write the calls down and give each one in turn a report. I can run through ten in one minute if I have the calls. And getting the calls shouldn't take more than one minute at best, right?

So on comes my helper with his list: "Hello, Juliet Yankee Eight Alpha Alpha. Juliet Yankee Eight Alpha Alpha, this is Kilo Baker Zero Xray Quebec Quebec. This is Kilo Baker Zero Xray Quebec Quebec in Washington, Nebraska. That's doubleyou whiskey, aye America, ess Sweden, aich hotel, eye Indiana, enn Nancy, gee George, tee Toronto, oh Oregon, enn Nancy. The name here is Wilbur . . . that's doubleyou whiskey, eye Indiana, ell love, bee bravo, you Uruguay, are radio. You're certainly putting a good signal in here, Wayne, really poking the old S-meter around. I have your list for you. I sure hope everyone on the list is going to come up here and find us. Let me know if you're getting me all okay and I'll pass along the list. Juliet Yankee Eight Alpha Alpha, this is Kilo Bravo Zero Xray Queen Queen standing by."

I grunt out a terse "go ahead, quickly, please," stifling an oath. Good old Wilbur comes back, spelling out both of our calls again, apparently in the belief that I may have forgotten not only his call, but my own, too. Then he starts slowly spelling out each call, one at a time, spelling each one twice for me and standing by after each call has been sent, triumphantly.

Wilbur is able to take a one-minute task and escalate it to nerve-shattering lengths for all concerned, stretching it out for maybe ten minutes or more. By this time, the other two chaps who have been scrounging calls for me are on channel wanting to give me their lists and I suddenly have thirty stations to work instead of just ten.

Ten works well. Thirty doesn't. Before I can work my way down through even twenty of the chaps, I have some breakers wanting to know who is on there, how can they get into the contact, where is the list being made . . . plus an Italian calling CQ DX with a 40 over 9 signal. When I ask him to stand by, he gives me my report, spells out his name twice, then the name of his town twice (why do most Italian towns have over twenty letters?), and refuses to get off the frequency until I give him my location and agree to QSL.

It is unfortunate that many lists are run in a much slower style. Hopefully, some of the DX operators will read this or hear about it and get moving faster. I heard YK1AA on the other evening and some OK station was running a list for him. It took maybe five minutes for each damned contact . . . and all Rasheed did was swap signal reports and call letters. But it went on with excruciating slowness and drove me and a few hundred others up the wall. I didn't particularly need a contact with Syria; I just wanted an opportunity to say hello to Rasheed since I had visited him and operated from his station a few years back.

Try the live list system with perhaps three good operators, all with good signals and in separate parts of the country, taking calls for the DX station. If you can't clearly read off ten calls, spelling them out as you go for the DX op to copy, in less than one minute, then you need some practice. Once the DX op has the calls, 95% of the work is over . . . he can then stand by for one, get the report, give a report, and get an okay . . . stand by for the next, etc. He doesn't have to spell any calls out or even give his own call more than once every few minutes.

The purpose of most DXpeditions is to work as many stations as possible in a given time . . . so why horse around with slow and frustrating systems?

MAY WINNER

"The W7GAQ Key Collection" collected an extra \$100 for author Martin Krey K7NZ, since his article was voted the best in our May issue. Make sure your favorite article gets recognized by using your Reader Service card ballot.

Microcomputer Interfacing

from page 26

unique digit enable code, and then output these eight bits of information to the interface. For instance, the value 01010011 will cause a five to be displayed on the fourth digit from the right on the display. The digit enable code in this example is 0011, and the digit enable code for the right-hand display is 0000.

To display the number 3,163,908,332, register pair H is first loaded with the memory address where the two least-significant digits (LSDs, 32) are stored. In Fig. 4, this address is 004 120. Only five memory locations are required to store the 10-digit number, because the number is stored in a packed BCD format. The D register, which is used to store the digit enable code, is then set to zero. The content of register D, which will eventually be latched by the interface hardware, is used to turn on one and only one digit at a time. At DISPL1, the DIGIT subroutine is called, so that the content of memory addressed by register pair H is moved to

the A register. Since the data values are stored in memory as packed BCD digits, this subroutine unpacks them. The digit in the four LSBs of the A register is rotated to the left four times, bit by bit. The first digit to be displayed is now in bits D7 through D4 of the A register when the OUTIT subroutine is called.

At OUTIT, the four LSBs of the A register are set to zero by the ANI instruction (the digit to be displayed is in bits D7 through D4). The digit enable code contained in the D register is then added to the BCD number in the A register and the result is output to the interface (OUT 125). When this word is output, the four LSBs determine which digit is enabled and the four MSBs represent the value to be displayed. The first time the OUTIT subroutine is executed, the A register will contain 00100000, because a two must be displayed on the right-hand digit.

At the end of the OUTIT subroutine, the content of the D register (the digit enable code)

/THIS PROGRAM DRIVES A 10-DIGIT, MULTIPLEXED,
/LIGHT EMITTING DIODE (LED), SEVEN-SEGMENT DISPLAY
/IN ADDITION, INSTRUCTIONS HAVE BEEN ADDED SO THAT
/THE INTENSITY OF EACH DISPLAY IS EQUAL.

```

DISPLA, LXIH      /LOAD REGISTER PAIR H WITH THE MEMORY
120      /ADDRESS WHERE THE BCD DIGITS ARE STORED.
004      /004 120 = HEX 0450
MVID      /LOAD D WITH THE FIRST DIGIT
000      /THAT WILL BE ENABLED
DISPL1, CALL      /DISPLAY THE FIRST TWO PACKED
DIGIT      /BCD DIGITS
0
INXH      /INCREMENT THE MEMORY ADDRESS
MOVAD      /GET THE DIGIT ENABLE WORD INTO A
CPI       /COMPARE IT TO THE
012      /ELEVENTH DIGIT ENABLE COUNT
JNZ       /HAVEN'T DISPLAYED ALL TEI
DISPL1      /DIGITS YET, SO DO TWO MORE
0
JMP       /HAVE DISPLAYED ALL 10 DIGITS,
DISPLA      /SO DISPLAY THEM ALL AGAIN
0
DIGIT, MOVAM      /GET THE PACKED BCD WORD INTO A
RLC       /ROTATE THE FOUR LSB BITS INTO THE
RLC       /FOUR MSB BITS
RLC
CALL      /THEN DISPLAY THIS DIGIT
OUTIT      /THEIR DISPLAY THIS DIGIT
0
MOVAM      /GET THE SAME WORD AGAIN
ANI       /SAVE ONLY THE FOUR MSBS
360      /(360 = HEX F0)
ADDI      /ADD THE DIGIT ENABLE
OUT       /OUTPUT THE EIGHT-BIT VALUE.
125
INRD      /INCREMENT THE DIGITENABLE
MVIE      /LOAD E WITH A NUMBER
200      /200 = HEX 80 = DECIMAL 128
INTEN1, DCRE      /DECREMENT THE NUMBER
JNZ       /IF IT IS NON-ZERO, EXECUTE THE
INTEN1      /JNZ INSTRUCTION BACK TO INTEN1
0
RET       /WHEN E = 0, RETURN
BINSS, 077      /SEVEN-SEGMENT CODE FOR 0
006      /SEVEN-SEGMENT CODE FOR 1
133      /SEVEN-SEGMENT CODE FOR 2
117      /SEVEN-SEGMENT CODE FOR 3
146      /SEVEN-SEGMENT CODE FOR 4
155      /SEVEN-SEGMENT CODE FOR 5
174      /SEVEN-SEGMENT CODE FOR 6
007      /SEVEN-SEGMENT CODE FOR 7
177      /SEVEN-SEGMENT CODE FOR 8
147      /SEVEN-SEGMENT CODE FOR 9
000      /SEVEN-SEGMENT CODE FOR 10
000      /SEVEN-SEGMENT CODE FOR 11
000      /SEVEN-SEGMENT CODE FOR 12
000      /SEVEN-SEGMENT CODE FOR 13
000      /SEVEN-SEGMENT CODE FOR 14
000      /SEVEN-SEGMENT CODE FOR 15

```

Fig. 4. Displaying a 10-digit number on a multiplexed LED display.

is incremented from zero to one, preparing it for the next digit enable code. To "intensify" the digit that has just been turned on, the computer executes a time delay at INTENS. Thus the digit is kept on for a period that allows it to be clearly seen. When the 8080 finishes the OUTIT subroutine, it returns to the MOVAM instruction just before the OUTIT subroutine.

The A register is loaded with the content of the same memory location addressed by register pair H, only this time the BCD digit contained in bits D7 through D4 must be displayed. This is the second digit from the right in the number 3,163,908,332. The ANI instruction at OUTIT sets bits D4 through D0 to zero and the digit enable code (one) is added to

/THIS PROGRAM DRIVES A 10-DIGIT, MULTIPLEXED, SEVEN-
/SEGMENT DISPLAY. A LOOK-UP TABLE IS USED TO CONVERT
/BCD NUMBERS TO THE APPROPRIATE SEVEN-SEGMENT CODE.

```

DISPLA, LXIH      /LOAD REGISTER PAIR H WITH THE MEMORY
120      /ADDRESS WHERE THE BCD DIGITS ARE STORED.
004      /004 120 = HEX 0450
MVID      /LOAD D WITH THE FIRST DIGIT
000      /THAT WILL BE ENABLED
DISPL1, CALL      /DISPLAY THE FIRST TWO PACKED
DIGIT      /BCD DIGITS
0
INXH      /INCREMENT THE MEMORY ADDRESS
MOVAD      /GET THE DIGIT ENABLE WORD INTO A
CPI       /COMPARE IT TO THE
012      /ELEVENTH DIGIT ENABLE COUNT
JNZ       /HAVEN'T DISPLAYED ALL TEI
DISPL1      /DIGITS YET, SO DO TWO MORE
0
JMP       /HAVE DISPLAYED ALL 10 DIGITS,
DISPLA      /SO DISPLAY THEM ALL AGAIN
0
DIGIT, MOVAM      /GET THE PACKED BCD WORD INTO A
CALL      /THEN USE THE LOOK-UP TABLE TO
OUTIT      /DETERMINE THE PROPER SEQUENCE
0      /OF ONES AND ZEROS.
MOVAM      /GET THE SAME WORD AGAIN
RLC       /ROTATE THE FOUR MSB BITS INTO
RLC       /THE FOUR LSB BITS.
RLC
OUTIT, ANI       /SAVE ONLY THE FOUR LSBs
017      /(017 = HEX 0F)
PUSHH      /SAVE REGISTER PAIR H ON THE STACK
LXIH      /LOAD REGISTER PAIR H WITH THE
BINSS      /BASE ADDRESS OF THE LOOK-UP TABLE.
0
ADDL      /ADD THE LO ADDRESS TO THE NUMBER IN A
MOVLA      /SAVE THE RESULT IN THE L REGISTER
JNC       /THERE IS NO CARRY, SO LEAVE THE
OKASIS      /CONTENT OF THE H REGISTER ALONE
0
INRH      /INCREMENT THE H REGISTER BY ONE
MVIA      /OUTPUT AN INVALID DIGIT ENABLE CODE
377      /SO THAT THE DISPLAY IS BLANCHED.
OUT      /125
MOVAM      /GET THE SEVEN-SEGMENT CODE INTO A.
POPH      /POP REGISTER PAIR H OFF OF THE STACK
OUT      /THEN OUTPUT THE VALUE TO THE
126      /INTERFACE (7475 AND UD12981).
MOVAD      /GET THE DIGIT ENABLE CODE
OUT      /AND OUTPUT IT TO THE INTERFACE
125
INRD      /INCREMENT THE DIGIT ENABLE
MVIE      /LOAD E WITH A NUMBER
200      /200 = HEX 80 = DECIMAL 128
INTEN1, DCRE      /DECREMENT THE NUMBER
JNZ       /IF IT IS NON-ZERO, EXECUTE THE
INTEN1      /JNZ INSTRUCTION BACK TO INTEN1
0
RET       /WHEN E = 0, RETURN
BINSS, 077      /SEVEN-SEGMENT CODE FOR 0
006      /SEVEN-SEGMENT CODE FOR 1
133      /SEVEN-SEGMENT CODE FOR 2
117      /SEVEN-SEGMENT CODE FOR 3
146      /SEVEN-SEGMENT CODE FOR 4
155      /SEVEN-SEGMENT CODE FOR 5
174      /SEVEN-SEGMENT CODE FOR 6
007      /SEVEN-SEGMENT CODE FOR 7
177      /SEVEN-SEGMENT CODE FOR 8
147      /SEVEN-SEGMENT CODE FOR 9
000      /SEVEN-SEGMENT CODE FOR 10
000      /SEVEN-SEGMENT CODE FOR 11
000      /SEVEN-SEGMENT CODE FOR 12
000      /SEVEN-SEGMENT CODE FOR 13
000      /SEVEN-SEGMENT CODE FOR 14
000      /SEVEN-SEGMENT CODE FOR 15

```

Fig. 5. Using a look-up table with the multiplexed LED display.

the number and the result output. After incrementing the digit enable code to two, the 8080 again executes the INTENS delay loop, so that the three on the second digit from the right is displayed for a reasonable amount of time. When the RET instruction is executed, the 8080 returns to the INXH instruction just before DISPL1.

The 8080 increments the memory address in register pair H and then examines the digit enable code contained in the D register. If this code is less than 012(decimal 10), the 8080 jumps back to DISPL1 so that the two BCD digits now addressed by register pair H are displayed. If the digit enable code is equal to 012, then all ten digits in the number have been displayed. If this is the case, the 8080 jumps back to DISPLA so that the memory address in register pair H and the digit enable code in the D register are re-initialized.

As you can see from this program, the 8080 is constantly updating (writing new information out to the interface) the display. New digit enable codes and data values are output to the display every millisecond.

Unfortunately, this "ties up" the microcomputer and it cannot perform any other tasks. Of course, the program listed in Fig. 4 could be converted to a subroutine. If it is called every 15 or 20 ms, the 10-digit number will still be displayed at a reasonably fast rate. However, it may be difficult to program the microcomputer so that the DISPLA subroutine is called this often.

One solution to this problem is to use an interrupt. This means that whenever the 8080 is interrupted, it outputs a numeric value to the next consecutive digit in the display. The microcomputer will have to be interrupted at least 400 times every second for the display not to "flicker." This means that our 10-digit number will be displayed 40 times every second. To interrupt the 8080 so that it "services" the display this frequently, a low frequency oscillator (400 Hz) can be wired to one of the interrupt interfaces previously discussed.^{2,3}

The program that services the interrupt and outputs a new

eight-bit value to the display interface is very similar to the software in Fig. 3. Because of its length, we have not included a listing of it. The only difference between the two programs is that the 8080 has to store the digit enable code and the address of the memory location that contains the next digit to be displayed. These values must be stored in memory, because register pair H and the D register may be used by the program that was interrupted. Therefore, each time the oscillator interrupts the microcomputer, it has to read from memory the address for the next digit to be displayed and its corresponding digit enable code. After the number is displayed, the digit enable code and the memory address have to be incremented and stored back in memory. To keep this software as simple as possible, it is easiest to store the 10-digit number in ten memory locations, one digit per memory location, rather than "packing" the BCD data.

One feature of our interface that we have not mentioned is the fact that we are limited by the DS8857 to display only the numbers 0 through 9. This device cannot be used if we wish to display the letters A through F, as is often the case if hexadecimal numbers or message codes need to be displayed. To display the hexadecimal characters 0-9 and A-F, the DS8857 must be replaced by a device that can be used to drive individual segments in any combination we desire. The UDN-2981A (Sprague Electric Company, Worcester MA) contains eight individual drivers that can be used in the display interface. This device has eight inputs and eight outputs—each input is interfaced to an individual data bus line by means of a latch circuit and each output is wired through a current-limiting resistor to one of the segment inputs of the display (A-G). If desired, the last driver in the integrated circuit can be used to drive the decimal point in the display. Since the UDN-2981A has eight inputs, an eight-bit output port (latch) will have to be used to interface it to the microcomputer. This means

that one output port will be used to output the sequence of ones and zeroes that specifies a specific sequence of segments to be turned on or off. A four-bit output port is also required so that the digit enable code can be output and stored in the interface.

In Fig. 5, there is a look-up table used for converting BCD numbers to seven-segment codes.

The first difference between the programs listed in Figs. 4 and 5 can be seen at DIGIT. After the packed BCD word is moved from memory to the A register, the OUTIT subroutine is called. This means that the first BCD digit to be displayed is in bits D3 through D0 of the A register. At OUTIT, bits D7 through D4 of the A register are set to zero. The content of register pair H is then saved on the stack and register pair H is loaded with the base address of the look-up table. The BCD number in the A register is added to this address. The MVIA and OUT instructions at OKASIS cause the display to be blanked by writing an invalid digit enable code out to the display interface. While the display is blanked, the 8080 moves the bit pattern for the number to be displayed from memory to the A register (MOVAM). Register pair H is then popped off the stack and the seven-segment bit pattern is output to output port 126. The A register is then loaded with the digit enable code which is output to output port 125. When the OUT 125 instruction is executed, the specified digit in the display is turned on. The 8080 executes an intensifying time delay loop before returning to the MOVAM instruction just after DIGIT.

The MOVAM instruction loads the A register with the same packed BCD word that contained the previously displayed digit. However, the BCD digit in bits D7 through D4 must now be displayed. Therefore, bits D7 through D4 are rotated into bits D3 through D0 of the A register before the 8080 executes the instructions at OUTIT which cause the required bit pattern for the seven-segment display to be fetched from the look-up table. The bit

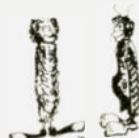
pattern followed by a digit enable code is then output to the interface. The remaining instructions in Fig. 5 are the same as the instructions in Fig. 4.

One advantage of using seven or eight individual segment drivers is that you can display any combination of segments that you want. You are no longer limited to the numbers 0-9. If desired, you can display the letters A-F so that hexadecimal numbers can be displayed. For some special applications, some additional letters and words can be displayed, such as H, r, L, o, OIL, HELP, Error, etc.

There is one additional display method that we have not discussed—the use of an external display controller integrated circuit to control the multiplexed display. The Intel Corporation makes a number of these integrated circuits that are compatible with the 4004/4040, 8080, and 8085. These are the 4269, 8279, and 8279-5 integrated circuits. National Semiconductor Corporation also has two display controller integrated circuits that can be used with six-digit displays. One of the devices (MM74C912) can be used to display 0-9 and the other (MM74C917) can be used to display hexadecimal numbers. For additional hardware and software information about multiplexed LED displays, we refer you to reference 4.

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Ham Help

I need an operator's manual and schematic for a BC-1031 panoramic adapter. I will gladly pay for all copying and postage.

Chuck Holstein WD5CUG
4721 Lucy Dr.
El Paso TX 79924

I would like to copy or purchase the manual and/or schematic for the Multi-Elmac AF-67 transmitter and power supply. Can anyone help? Thanks.

Al Christman WD8CBJ
Box 44
Granville WV 26534

I have some older VHF equipment that utilizes electron tubes (remember those things?). Anyway, there doesn't appear to be a ready market for the tubes I need and I can't find a distributor who sells them.

If anyone has one or more tubes I list below, I would appreciate hearing from them, particularly those having the 6850. The tubes I need are: 5696, 5718, 5896, 5899, 5902,

5940, 6021, 6112, and 6850. Many thanks.

Bill Walston KB5GD
Drawer 1418
Rockport TX 78382

Does anyone have a Geiger counter in good working condition reasonably priced? Please send me the particulars.

Leroy Lawmaster W5HOM
Rt. 1 Oak Hill
Westville OK 74965

DX

from page 12

clerk.

Although the story seems to be changing daily, it now appears that the DXCC desk will not accept TH8JM cards for credit. If you need this one, you will probably have to work John again when he gets his true-blue TL8 license.

TR8AC and TR8GDC recently moved to Alexandria, Egypt, and should be heard signing SU any day.

Anyone needing confirmation for an ET3FMA/9E3FMA/9F3FMA contact from 1965 to 1968 can send an SASE to Don Murray W4WJ, 19700 NW 5 Ct., Miami FL 33169.

ON4UN recently worked CR9AJ on 75 meters to complete his 5BWAZ. Five more QSLs and it will be confirmed. 5BWAZ is easily the most difficult operating award to obtain in all of hamdom.

The correct route for ZS2MI QSLs is WA2IEN, 255 Route 17, Upper Saddle River NJ 07458.

C5ABK has reportedly received permission to operate in CR3 Guinea-Bissau for two weeks in December. He is presently assembling the necessary equipment to make the operation a success and plans to open up on December 1st. QSL to G3LQP.

If you worked HV3SJ on November 11/12 and had your QSL bounced by W6KNH, try again. Logs for that period disappeared for awhile, but now they are safely in the hands of W6KNH, all 45 pages.

As we mentioned last month, several of the deserving were surprised last June when they received a citation from the FCC for contacting HS stations in Thailand. The citation stated that Thailand was on the banned countries list. Those responding received a follow-up form letter from the FCC saying only that in view of the reply, the matter was being closed. Apparently, no form letter exists for apologizing and admitting an error.

The Panamanian government has confirmed to the ARRL that HP9 callsigns have never been issued for use on amateur bands. Communication of any type with stations signing HP9 is illegal.

VR6TC has recently acquired some SSTV gear and has been showing regularly on 14233 kHz from 0530Z. QSL to W6HS.

All KZ5 Canal Zone licenses expire September 30, 1979.

9M2FK reports needing Arkansas, Utah, North Dakota, and South Dakota for WAS and

Zone 2 for WAZ. Look for him on 14020 kHz between 1000Z and 1300Z.

Those of you involved in the DX net and/or DX list business should keep in mind that it is possible to violate third-party traffic regulations when passing a list to a DX station. If you compile a list and pass it directly to a station in a non-third-party country, then you could be in violation. If you say "DX1DX call W5USA," then you are in violation. The correct way would be to say "W5USA call DX1DX." If you think this is just nit-picking, then be advised that some FCC monitoring types are on the prowl for just such violations.

Jim Smith P29JS finally received an answer from the Indian government concerning his application for permission to operate from Andaman. Unfortunately, the letter stated that new rules beginning in January, 1979, made it impossible to grant licenses for short stays.

Rumors floated hot and heavy that XZ2P was SP5AUC and that his was a true-blue operation. Unfortunately, the real SP5AUC denied the rumor and that sort of ended that.

While in Poland, we might mention that Polish amateurs are now allowed to work 160 meters between 1750 kHz and 1950 kHz. Maximum power is 10 Watts.

New officers of Murphy's Marauders include Al Meleg N1JW—President, Ron Grzelak K1BW—Vice President, Dan Street K1TO—Activities Manager, and Ron Nevers K1TVM—Secretary/Treasurer. Newsletter editor is Ed Kalin K1RT.

WA1SQB ran off better than 5000 contacts during his HC8 operation in March. He should

be setting up in Montserrat this month where he will be signing either VP2MAY or VP2MI.

Last month we mentioned that sending currency through the mails to foreign hams could bring trouble to the recipient. If you are in doubt as to which countries can receive USA currency in the mails, ask your local post office to check through the Foreign Postal Manual for the country in question. Any illegal items are listed in this manual.

If you've worked everything else, then you might try for all the cities and districts in Japan. A full list can be obtained by sending 14 IRCs or \$8.00 to Michio Koshimizu JR1BFT, 37-3, Nakano, 1-chome, Nakano-ku, Tokyo 164, Japan.

Tom Wong VE7BC recently returned from a business trip to China. While he was there, he inquired into the possibility of obtaining an operating permit for a future visit. Tom says that the current policy of the Central Government is that there will be no amateur radio operation. It seems that there will probably have to be some top-level changes made before any legitimate operation occurs. Tom feels, however, that this is entirely possible, and with the great interest in amateur radio, something might develop before the end of the year. The feeling still runs strong that when a legitimate operation does surface, it will be by the Chinese themselves and not a group of war-mongering, Imperialist, running-dog lackeys, or whomever. Keep the faith.

That Soviet ski team that was signing U0CR successfully reached the North Pole last July.

ZS5DC is Diane Cardell, formerly VQ9DC, now living in Durban. Diane was in on the big Desroches effort by VQ9D, VQ9BP, and others back in 1974.

That HB9APN/BY operation a

few months back did not have official approval and is not being accepted for DXCC credit.

Several "goodies" have been checking into the WA2JUQ Net on 14240 kHz. Heard recently were TA1MB, YI1BGD, FH8YL, and ZS2MI on Marion Island. The time to listen is around 0630Z.

Another net to watch is the DX-DX Net on 21280 kHz. Listen from 1600Z.

Vendaland is the third of the nine African homelands being established by South Africa, the first two being S8 Transkei and H5 Bophuthatswana. Independence day is set for September 13 and several ZS types plan to be there for the unveiling. ZS6AK and ZS6ABO will head a contingent of eight Durban DXers planning to make the 200-mile trip. DXCC status is expected to be the same as Transkei and Bophuthatswana, that is, no decision until after WARC 79 and then a favorable decision made retroactive back to the date of independence.

JF1IST/7J showed on schedule from Okino Torishima with a very good signal to most of the states. The one-man effort opened up on June 11th and ran through June 15th with WB8LDH the last station logged at 0653Z. Weather conditions cut short the expected ten-day stay, but some 5500 contacts were made on 6 through 40 meters. QSL to JA1RNH, Itaru Tomita, 1-2-9, Zaimokuza, Kamakura, 248 Japan.

This is the month (September) the ARRL begins accepting Desecheo QSLs for DXCC credit.

Reports have all 7Q7 stations being shut down and their equipment confiscated.

Slim showed from 3Y6CD in June and from 9A1VU earlier in the year.

The correct route for all FH8OM/YL QSLs is to Box 86, Dzazoudi, 97600 Mayotte . . . via France.

The Guatemalan Radio Society indicates that TG7AA was Slim and the ARRL is returning the cards without allowing DXCC credit.

There is a multi-national effort being planned for a fifteen-day operation from 3V8 Tunisia this month. Amateurs from France, Italy, and the UK will man as many as ten different stations.

UHF marine-band applicants are being asked to make up their own callsigns until the overworked FCC can get their applications processed and the regular type licenses in the mail.

If you paid more than \$20.00 to the FCC between August 1, 1970, and December 31, 1976, you are eligible for a refund. If you are eligible, then you



Jean F5FV and his nice layout in France. When not chasing DX for himself, Jean serves as QSL Manager for FB8XV.

should obtain a copy of the "Phase One Fee Refund Program" instruction and form manual from your nearest FCC field office or Federal Information Center. Check the June QST for more information.

Remember that EI8H/anything is always Slim and you will save time and money.

Looking West

from page 10

USER RIGHTS

Question: What rights, if any, do repeater users have when being repeated through the station of another amateur? This is a question being fiercely debated in southern California and elsewhere. It recently surfaced here in Los Angeles when one of the control operators on the WA6KOS/R system turned the repeater off on a system user because of the "content of communications." The operator in question yelled "censorship" and the controversy ensued.

One group feels that the system owner, licensee, or control station does not have the right to turn off the repeater for any reason other than a technical malfunction, and that the repeater is akin to a public utility and therefore must be kept on the air regardless of what is repeated. This group feels that the total responsibility for "content of communications" rests with the individual system user, and the use of jamming, music, and even profanity is left to his discretion. To turn a repeater off under these circumstances is an act of censorship against the originating station.

The other side yells "wrong!" The FCC rules clearly state that an amateur station may not transmit profanity, unidentified signals, or music, and that a repeater does constitute an amateur station. Therefore, not only do the system licensee and control operators have a right to terminate system operation in such circumstances as outlined above, but under the terms of Parts 97.113, 97.115, 97.116, 97.119, 97.123, and 97.125 of the FCC Rules and Regulations, they are directly *obligated* to take such action. To do otherwise directly violates FCC rules. This group also feels that once an amateur decides to operate through a repeater, he is, in effect, operating through the station of another amateur. Therefore, when operating through a repeater, the user gives up any First Amendment rights and is totally subject to the will of the system owner/licensee. What-ever standards, guidelines, and

Bill KA2WG writes to remind us all that even though KA stations are not recognized as amateur stations by the Japanese government, they still count as Japan for everything except the All Asia Contest. Any US ham being transferred to Japan or Okinawa as US military or a civilian employee of the

military is eligible to obtain a KA license. For more information, contact FEARL, c/o Sam Flemming, USAGH D/O, APO SF 96343.

That's about it for this month. I hope some of the information provided here will help you pick up a few new ones. If it does, let me know. In the meantime,

don't forget we are always looking for DX info and pictures. Any pictures you send us can be returned if you wish. In the meantime, best DX and health.

Thanks as always to the West Coast DX Bulletin, LIDXA Newsletter, and WorldRadio Magazine for much of the preceding information.

regulations the licensee sets for user operation must be adhered to without question.

Where do we find the answers? Part 97? Have any of you read Part 97 lately? The FCC regulations, in relation to repeater operation, are too ambiguous. For that reason, I have written a letter to the FCC asking direct questions about repeater operation, regulatory enforcement, and other related matters. The letter was written on June 19, 1979, and was mailed to the FCC on June 23rd. It is presented below for your information. If received, the FCC's response will also be presented in this column.

Federal Communications
Commission
M Street
Washington DC
Attn: Personal Communications
Division

Gentlemen:

I wish to ask some specific questions in regard to Part 97 of the Amateur Rules and Regulations, concerning amateur repeater operation.

- 1) Who is responsible for "content of communications," the system licensee or the system user?
- 2) If a repeater system owner permits profanity and unidentified transmissions to be "repeated" via his repeater system (while his system is operational under fully automatic remote control), is he legally responsible for the content of such communications? Can he legally censor such communications? Should he?
- 3) Does the FCC hold any one person legally responsible for "content of communications" and all other aspects of amateur repeater operation?
- 4) What constitutes an operational amateur repeater—an established repeater, known to the community, or a repeater system that happens to be operating at any given moment?
- 5) If an established amateur repeater is resting between transmissions and another amateur decides to operate during that time, is the established amateur guilty of jamming when he operates once again, or is the new operator considered responsible for the subsequent jamming?
- 6) If the conduct of any amateur is considered by his peers to be obnoxious, although he may be technically operating within the regulations of Part 97, can the amateur community, through the Commission, take any legal action to stop said amateur?
- 7) Why is the Commission unwilling to heed the calls for help from amateurs

who are suffering the outrage of willful and malicious interference on both high frequency SSB and VHF FM repeater operations?

8) Is the Commission willing to enforce the terms of Part 97 of the Amateur Rules and Regulations regarding profanity and unidentified transmissions (97.116, 97.119, and 97.123)? If not, why not?

9) Why hasn't the Commission acted to suspend or revoke the amateur license of one Scott Lookholder WB6LHB? He was convicted in federal court on a charge of using foul and abusive language on the two-meter amateur band. I have copies of letters sent to the Commission by southern California amateurs requesting such action. Does the Commission intend to act in this matter?

10) What action can we amateurs take to rid the community of the individuals who, licensed or unlicensed, willfully and maliciously jam, harass, and intimidate the law-abiding amateur's day-to-day operation? Would the Commission be willing to act if supplied documentation of such violations? Exactly what type of documentation does the Commission require before acting on such reported violations?

11) If I challenge the right of another amateur to hold an amateur's license because he operates his amateur station in violation of the regulations as set forth in Part 97, what possible Commission action can be taken? Must I file a formal challenge with your office?

12) Please define the following (using specific examples) as they relate to our subject:

- A) willful and malicious interference
- B) unidentified transmissions
- C) profanity and indecency

I intend to publish this letter—and your response—in my "Looking West" column. Many amateur repeaters are suffering tremendous willful and malicious interference problems, and it is my hope to help guide them toward a solution. Your response to these questions will be appreciated.

Yours truly,
William M. Pasternak

I wish to credit WB2MIC for arousing my interest in this subject. In the June issue of 73, a letter appeared from Jozef which relates directly to the situation out here. Jozef felt that he was treated unfairly by a repeater licensee who may have arbitrarily censored him. I must state my personal belief that, while I may not agree with arbitrary censorship by control operators and system licensees in certain situations, these people are clearly within their rights when they choose to censor. While you and I may not want to adhere to these standards, we are obligated to do so because we really have no alternative. I remember being told by one

amateur that he felt he was a "god" by virtue of the fact that he owned and operated his own repeater. If he demanded that his users "paint their radios with red stripes and operate only while standing on their heads," they had better adhere to his dictates "or else." He had absolutely no compunction whatsoever about turning off his repeater if someone even "sneezed the wrong way." He was clearly "the boss."

It sounds absurd, but that's the way he felt. Having never operated through his system, I cannot tell you how tight a ship he really runs. Jozef and many others are unhappy with this unresolved situation. The problem is spreading. Only the FCC can make a final determination of this matter, and we await their answer.

HANDLING MALICIOUS INTERFERENCE DEPARTMENT

A few months ago, "Looking West" ran a series of articles on the problem of how to deal with malicious interference. While the amount of response to my questions has not been overwhelming, I have received a few letters. Over the next few months, I hope to bring you some of my respondents' views.

The following letter appeared in the June issue of the Mt. Lee (CA) Repeater Association Newsletter and is relevant to our discussion:

The jammers must be getting quite a laugh out of repeater users and owners these days, because a lot of us are so easy to bait.

A jammer with nothing better to do will figure out someone's weakness and rub that person the wrong way. Often this jamming is legal, because the one doing the jamming is a ham and uses proper procedure. The jammer, by manipulating the right people with the right words, can tie up a repeater with hours of counter-jamming, shouting matches, and arguments. Former "good guy" hams, when baited and sufficiently angered, may resort to counter-jamming the original troublemaker. The jammer's gotten what he wants: People have taken his bait. A good example of this might be the licensed ham who uses CW on a repeater. He uses correct procedure, callsigns, and breaks properly, but may be trying to bait a response. CW, RTTY, and SSTV are legal on repeaters if mode and bandwidth rules are obeyed. Instead of reacting with hostility, come back with perfect F1 30-wpm telegraphy! What rubs you the wrong way? Do you fall for the bait?

Some of us feel that high power is

the solution to jamming. Boy, it feels good to talk right over the twerp with your 150 Watts and listen to the capture effect on your duplex! Secretly, you probably hope the jammer also has duplex so that he can hear his puny signal get stomped. Although this is satisfying, it creates problems. With 150 Watts versus the 2 Watts of the jammer, the DF committee will have a hard time hearing the jammer. Many jammers don't care if they are Q-5 in a repeater or not, because they get their jollies causing discomfort to other hams.

"Sorry OM ... no copy ... you're being jammed ... 73." You, the legitimate ham, have just succumbed to a psychological defeat from the jammer. The jammer is laughing at you. Try to sign off gracefully, if you're not succeeding. Don't give him his kicks by going QSY or QRT. Try to draw

the jammer out the same way he baits us. Bait the jammer, legally ... don't talk to him, talk about him to another ham. Don't get into jamming him. Instead, go to low power and allow plenty of time for breaks. It's unpleasant to hear one of these clowns and you may want to "cover" him, but be sure to leave a space for the DF committee to listen in. While drawing out a jammer, it is important not to act as he does. It's easy to get mad and jam the jammer. Aren't we better than that?

Most of us have a local repeater ... a machine we like to hang out on, meet our friends, etc. It is hard to break away from a favorite machine and go somewhere else. We are loyal to the repeater or club to which we belong. When most of us get mad at a jammer, we want to make our stand on our own machine. When the jammer is forced off our repeater or decides to hassle

another group, we tend to become temporarily apathetic about our common problem.

Jammers have knobs on their radios, although it seems some hams don't ... and these hams should be helping the ham community by following the jammer around the dial. All repeater users should get some sense of community and try to pool personnel and equipment to beat this thing. We must be able to continue the hunt even if the jammers have left "our" repeater and moved to another. The jammers seem secure in knowing that they can change frequencies without other repeater groups following.

DFers already know what to do, and the average user can also help. When the carriers or cursing start, listen to the input to the repeater. Normally, your radio is set so that you are transmitting on 147.84 MHz (the input) and

receiving on 147.24 MHz (the output). Reverse this setup, or at least find a way to listen on 147.84. (Transmitting on 147.24 is not necessary.) Now, check your S-meter; is the needle 25% over from 0? 80%? 100%? Break with your call, QTH, cross-streets, time of day, and date, and tell the "woodwork" what % reading you have. See if any other breakers would like to give their reports. (We use % of signal instead of S-units because S-meters are not standard or accurate.)

Let's hope we can end apathy on the airwaves and pool our resources. After all, the jamming is only a by-product of our apathy. The problems on repeaters affect us all, and high-power amps, private systems, and 800-channel radios are not adequate solutions to our common problem.

73,

An Interested User

Review

TELEVISION HANDBOOK FOR THE AMATEUR

A large number of hams are professionals in the field of commercial television. Their daily involvement in the world of video gives them a headstart in understanding and operating a fast scan television station. The rest of the population shouldn't be discouraged, though. The book, *Television Handbook for the Amateur*, is a real help in bridging the gap between beginner and pro.

Television Handbook is published by Aptron Laboratories and a paperback edition costs

\$6.50. The author, Biagio Presti, is not a ham, but as president of Aptron, he is familiar with amateur television efforts. The first two-thirds of the book is devoted to theory, while the remaining 30 pages discuss construction projects and include tables dealing with radio transmission and television standards.

By minimizing the use of mathematics and including numerous illustrations, the author is able to make the theory chapters down to earth and very understandable. The circuits used to generate a video signal are thoroughly

described. To get the most benefit from these chapters, the reader should have a good background in basic transistor circuit design. If you are studying for an Extra class license, the *Television Handbook* may be helpful. Questions concerning television theory appear on some versions of the exam.

There is an abundance of information about special circuits that won't be found in the usual amateur references. Color television may be more of a dream than a reality for most beginning ATVers, but the reader's appetite is whetted by a short summary of the principles involved. The contents of these chapters will be of interest long after your first ATV QSO.

The construction chapters are similar to the theory section.

The sub circuits that comprise 450-MHz transmitters and receivers are tackled individually. It is not like the wire-by-wire description frequently found in some ham publications. There is adequate information for the experienced home-brewer. As a bonus, you will have a good understanding of the theory behind the design.

The recent appearance of commercial equipment and beginner-oriented literature has helped to put fast scan television within the reach of many hams. *Television Handbook for the Amateur* does not give complete coverage of the subject, but for a ham who aspires to be more than an appliance operator, it is bound to be useful.

Tim Daniel N8RK/1
Peterborough NH

OSCAR Orbits

Courtesy of AMSAT

The listed data tells you the time and place that OSCAR 7 and OSCAR 8 cross the equator in an ascending orbit for the first time each day. To calculate successive OSCAR 7 orbits, make a list of the first orbit number and the next twelve orbits for that day. List the time of the first orbit. Each successive orbit is 115 minutes later (two hours less five minutes). The chart gives the longitude of the day's first ascending (northbound) equatorial crossing. Add 29° for each succeeding orbit. When OSCAR is ascending on the other side of the world from you, it will descend over you. To find the equatorial descending longitude, subtract 166° from the ascending longitude. To find the time OSCAR 7 passes the North Pole, add 29 minutes to the time it passes the equator. You should be able to hear OSCAR 7 when it is within 45 degrees of you. The easiest way to determine if OSCAR is above the horizon (and thus within range) at your location is to take a globe and draw a circle with a radius of 2450 miles (4000 kilometers) from your QTH. If OSCAR passes above that circle, you should be able to hear it. If it passes right overhead, you should hear it for about 24 minutes total. OSCAR 7 will pass an imaginary line drawn from San Francisco to Norfolk about 12 minutes after passing the equator. Add about a minute for each 200 miles that you live north of this line. If OSCAR passes 15° east or west of you, add another minute; at 30°, three minutes; at 45°, ten minutes. Mode A: 145.85-95 MHz uplink, 29.4-29.5 MHz downlink, beacon at 29.502 MHz. Mode B: 432.125-175 MHz uplink, 145.975-925 MHz downlink, beacon at 145.972 MHz.

OSCAR 8 calculations are similar to those for OSCAR 7, with some important exceptions. Instead of making 13 orbits each day, OSCAR 8 makes 14 orbits during each 24-hour period. The orbital period of OSCAR 8 is therefore somewhat shorter: 103 minutes.

To calculate successive OSCAR 8 orbits, make a list of the first orbit number (from the OSCAR 8 chart) and the next thirteen orbits for that day. List the time of the first orbit. Each successive orbit is then 103 minutes later. The chart gives the longitude of the day's first ascending equatorial crossing. Add 26° for each succeeding orbit. To find the time OSCAR 8 passes the North Pole, add 26 minutes to the time it crosses the equator. OSCAR 8 will cross the imaginary San Francisco-to-Norfolk line about 11 minutes after crossing the equator. Mode A: 145.85-95 MHz uplink, 29.4-29.5 MHz downlink, beacon at 29.40 MHz. Mode J: 145.90-146.00 MHz uplink, 435.20-435.10 MHz downlink, beacon on 435.090 MHz.

Orbit	OSCAR 7 Orbital Information			Orbit	OSCAR 8 Orbital Information		
	Date (Sep)	Time (GMT)	Longitude of Eq. Crossing °W		Date (Sep)	Time (GMT)	Longitude of Eq. Crossing °W
21928	1	0053:46	78.5	7593Jbn	1	0056:12	60.0
21941	2	0148:03	92.1	7607Jbn	2	0101:22	61.3
21953qrp	3	0047:23	76.9	7621Abn	3	0106:31	62.6
21966	4	0141:40	90.5	7635Abn	4	0111:41	63.9
21978X	5	0041:00	75.4	7649X	5	0116:50	65.2
21991	6	0135:17	89.0	7663Abn	6	0122:00	66.5
22003	7	0034:38	73.8	7677Abn	7	0127:09	67.8
22016	8	0128:55	87.4	7691Jbn	8	0132:19	69.1
22028	9	0028:15	72.3	7705Jbn	9	0137:28	70.4
22041qrp	10	0122:32	85.9	7719Abn	10	0142:38	71.7
22053	11	0021:52	70.7	7732Abn	11	0004:33	47.2
22066X	12	0116:09	84.3	7746X	12	0009:43	48.5
22078	13	0015:29	69.1	7760Abn	13	0014:52	49.8
22091	14	0109:46	82.7	7774Abn	14	0020:02	51.2
22103	15	0009:06	67.6	7788Jbn	15	0025:11	52.5
22116	16	0103:23	81.2	7802Jbn	16	0030:20	53.8
22126qrp	17	0002:44	66.0	7816Abn	17	0035:30	55.1
22141	18	0057:00	79.6	7830Abn	18	0040:39	56.4
22154X	19	0151:17	93.2	7844X	19	0045:48	57.7
22166	20	0050:38	78.1	7858Abn	20	0050:58	59.0
22179	21	0144:55	91.7	7872Abn	21	0056:07	60.3
22191	22	0044:15	76.5	7886Jbn	22	0101:16	61.6
22204	23	0138:32	90.1	7900Jbn	23	0106:26	62.9
22216qrp	24	0037:52	75.0	7914Abn	24	0111:35	64.2
22229	25	0132:09	88.5	7928Abn	25	0116:44	65.5
22241X	26	0031:29	73.4	7942X	26	0121:53	66.8
22254	27	0125:46	87.0	7956Abn	27	0127:03	68.1
22266	28	0025:06	71.8	7970Abn	28	0132:12	69.4
22279	29	0119:23	85.4	7984Jbn	29	0137:21	70.7
22291	30	0018:44	70.3	7998Jbn	30	0142.30	72.0

Awards

from page 28

SOUTH AMERICA

CE	Chile
CE0A	Easter Is.
CE0X	San Felix
CE0Z	Juan Fernandez
CP	Bolivia
CX	Uruguay
FY	French Guiana
HC	Ecuador
HC8	Galapagos Is.
HK	Colombia
HK0	Bajo Nuevo
HK0	Malpelo Is.
HK0	San Andres & Providencia
HP	Panama
HR	Honduras
HR0	Swan Is.
KZ	Canal Zone
LU	Argentina
OA	Peru
PJ	Bonaire
PJ	Netherlands Antilles
PY	Brazil
PY0	Fernando de Noronha
PY0	St. Peter & St. Paul
PY0	Trinidad & Martin Vaz Is.
PZ	Surinam
TG	Guatemala
TI	Costa Rica
TI9	Cocos Is.
VP1	Belize
VP8	Falkland Is.
VP8, LU	South Georgia Is.
VP8, LU	South Orkney Is.
VP8, LU	South Sandwich Is.
VP8, LU	South Shetland Is.
VP8W	South Grahamland
YN	Nicaragua
YS	Salvador
YY	Venezuela
YY0	Aves Is.
ZP	Paraguay
ZR	Guyana
9Y	Trinidad and Tobago

EUROPE

C3	Andorra
CT	Portugal
CT2	Azores
DA-DL	Federal Republic of Germany
DM, DT	German Democratic Republic
EA	Spain
EA8	Balearic Islands
EI	Republic of Ireland
EJ0	Aran Is.
F	France
FC	Corsica
G	England
GD	Isle of Man
GI	Northern Ireland
GJ, GC	Jersey
GM	Scotland
GM	Orkney Islands
GM	Shetland Islands
GU, GC	Guernsey
GW	Wales
HA	Hungary
HB	Switzerland
HB0	Liechtenstein
HV	Vatican
I	Italy
IC	Ischia
IA	Tuscan Archipelago
IS	Sardinia
IT	Sicily
JW	Bear Is.
JW	Svalbard Is.
JX	Jan Mayen
LA	Norway
LX	Luxembourg
LZ	Bulgaria
M1	San Marino
OE	Austria
OH	Finland
OH0	Aland Is.
OJ0	Market Reef
OK	Czechoslovakia
ON	Belgium
OY	Faeroe Islands
OZ	Denmark
PA	Netherlands
SM	Sweden
SP	Poland
SV	Greece
SV	Crete
SV	Dodecanese
SV	Mount Athos
TF	Iceland
UA, UK1, 3, 4, 6	European RSFSR
UA1, UK1	Franz Josef Land
UA2, UK2F	Kaliningradsk
UB, UK, UT, UY5	Ukraine
UC2, UK2	White RSFSR
UO5, UK50	Moldavia
UP2, UK2B, P	Lithuania
UO2, UK2G, Q	Latvia
UR2, UK2R, T	Estonia
YO	Romania
YU	Yugoslavia
ZA	Albania
ZB	Gibraltar
3A	Monaco
4U	ITU, Geneva
9A	(See M1)

ASIA

A4X	Oman Is.
A5	Bhutan
A8X	United Arab Emirates
A7X	Qatar
A9X	Bahrain
AP	Pakistan
BV	Taiwan
BY	China
CR9	Macao
EP	Iran
HL, HM	North Korea
HL, HM	South Korea
HS	Thailand
HZ, TZ	Saudi Arabia
JA-JR	Japan
JR6, KA6	Okinawa (Ryukyu Is.)
JD, KA1	Ogasawara
JT	Mongolia
KA	US Military in Japan
OD	Lebanon
S2	Bangladesh
TA	Turkey
UA, UK, UV,	Asiatic RSFSR
UW9-0	
UD6, UK6C, D, K	Azerbaijan
UF6, UK6F, O	Georgia
O, V	
UG8, UK6G	Armenia
UH8, UK8H	Turkmen
UI8, UK8I	Uzbek
UJ8, UK8J, R	Tadzhik
ULT, UK7	Kazakh
UM8, UK8M, N	Kirghiz
VS6	Hong Kong
VS9K	Kamaran Is.
VU	India
VU7	Andaman & Nicobar
VU7	Laccadives
XU	Khmer Republic
XV	Vietnam
XW	Laos People's Dem. Rep.
XZ	Burma
YA	Afghanistan
YI	Iraq
YK	Syria
IS	Spratty
4S	Sri Lanka
4X, 4Z	Israel
SB4, ZC	Cyprus
824	Neutral Zone
9H	Saudi Arabia/Iraq
9H4	Malta
9K	Gozo & Comino
9M2	Kuwait
9M6	West Malaysia
9M8	North Borneo
9N	Sarawak
9V	Nepal
Abu Ali, Jabal Attair	Singapore

OCEANIA

A3	Tonga Republic
CR8	Portuguese Timor
C2	Republic of Nauru
DU	Philippines
FK	New Caledonia
FO	French Polynesia
FW	Walls & Fortuna Islands
H4, VR4	Solomon Islands
JD, KA1	Minami Torishima
JD, 7J1	Okinawa Torishima
KB, KH1	Baker, Howland, American
KC6	Phoenix
KC6	Eastern Carolines
KG6, KH2	Western Carolines
KG6R	Guam Island
KG6S	Rota
KG6T	Saipan
KH6	Tinian
KH7	Hawaiian Islands
KJ, KH3	Kure Island
KM, KH4	Johnston Island
KP6, KH5	Midway Island
KP6, KH5	Kingman Reef
KS6, KH8	Palmyra
KW, KH9	American Samoa
KX	Wake Island
P2	Marshall Islands
T2, VR8	Papua, New Guinea
VK	Tuvalu Island
VK	Australia
VK9	Lord Howe Island
VK9	Willis Island
VK9	Christmas Island
VK9	Keeling, Cocos Island
VK9	Mellish Reef
VK9	Norfolk Island
VK9	Macquarie Island
VR1	British Phoenix Islands
VR1	Gilbert Island
VR1	Ocean Island
VR3	Christmas Island
VR6	Pitcairn Island
VR7	Line Island, South
VR8	and Central (See T2)
VSS	(See T2)
YB, YC, YD	Brunel
YB, YC, YD	Borneo
YB, YC, YD	Celebes
YB, YC, YD	Java
YB, YC, YD	Sumatra
YB, YC, YD	West Irian
YJ	New Hebrides
ZK1	North Cook Island
ZK1	South Cook Island

ZK2	Niue Island
ZL	New Zealand
ZL	Auckland & Campbell
ZL	Chatham Island
ZL	Kermadec
ZM7	Tokelau
3D2	Fiji Islands
5W	Western Samos

A2	Botswana
C5	Gambia
C9	Mozambique
CN	Morocco
CN2	Tanger
CR3	Guinea Bissau
CT3	Madeira Is.
D2, 3	Angola
D4	Republic of Cape Verde
D6	Comoros
EA8	Canary Islands
EA9	Ceuta and Melilla
Ifni	
EA9	Rio de Oro
EL	Liberia
ET2	Eritrea
ET3	Ethiopia
FBBW	Crozet
FBBX	Kerguelen Is.
FBBZ	Amsterdam & St. Paul
FH	Mayotte
FR	Glorioso Island
FR	Juan de Nova, Europa
FR	Reunion
FR	Tromelin
FR	Bophuthatswana
FR	Lampedusa Island
FR	Pantelleria Island
J2, FL8	Djibouti
S7	Seychelles
S8	Transkei
S9	Sao Tome and Principe
ST	Sudan
ST0	South Sudan
SU	Egypt
TJ	Cameroon
TL	Central African Empire
TN	Congo
TR	Gabon

TT	Ivory Coast
TU	Benin
TZ	Mali
VK8	Heard Island
VQ9	Alabdra Island
VQ9	Chagos (Diego Garcia)
VQ9	Desroches
VQ9	Farquhar
XT	Upper Volta
ZD7	St. Helena
ZD8	Ascension Island
ZD9	Gough Island and Tristan da Cunha

ZS1, 2, 4, 6	South Africa
ZS2	Prince Edward Island
ZS2	Marion Island
ZS3	Southwest Africa (Namibia)
ZE	Agalega & St. Brandon
ZE	Mauritius
ZE	Rodriguez Island
ZE	Equatorial Guinea
ZE	Swaziland
ZE	Tunisia
ZE	Republic of Guinea
ZE	Bouvet Island
ZE	Yemen
ZE	Libya
ZE	Tanzania
ZE	Nigeria
ZE	Malagasy Republic
ZE	Mauritania
ZE	Niger
ZE	Togo
ZE	Uganda
ZE	Kenya
ZE	Somali
ZE	Senegal
ZE	People's Dem. Rep. of Yemen
ZE	Lesotho
ZE	Malawi
ZE	Algeria
ZE	Maldives Islands
ZE	Ghana
ZE	Zambia
ZE	Sierra Leone
ZE	Republic of Zaire
ZE	Burundi
ZE	Rwanda

Corrections

In "So You Want to Raise a Tower" (July, 1979, page 110), Photo J shows what appears to be a vertical stacking of the cable lengths under the cable clamps. The cable lengths should be positioned in a horizontal side-by-side arrangement such that both halves of the cable clamps contact each side of each cable.

Gene Smart WB6TOV/1
News Editor

In reference to "At Last! A Really Simple Speech Processor," June, 1979:

Thanks to W5VSR for pointing out the need for a circuit change. A 10k pot should be added at the emitter of the first transistor.

Sorry about any inconvenience this may have caused.

Jeff Stadelman W9UT
Pound WI

Ham Help

I just acquired an 80-10-meter amp. It's a TXL-80B made by Westcom Engineering. The problem is that the final transistors have been blown and I can't find any suitable replacements, as the transistors have a nonstandard "house number." I'd be grateful to anyone who could suggest a good replacement, or give me an address where I might find Westcom. Thanks in advance for helping.

Bob Howle WA4ZID/0
10432 Baseline Road
Lafayette CO 80026

I'm the president of the newly-formed Phoenix Honeywell Amateur Radio Club and could use some help. We have a lot of tube-type equipment and most of us would like to replace the front-end rf stage with an

FET equivalent (such as a 6HA5). We would like to build them on a direct-replacement tube socket. Does anyone have tubes-to-FET conversion circuits? Also, does anyone have any application circuits for FET noise blankers or audio noise filters? Any help will be appreciated.

P.J. Scola, President
Phoenix Honeywell Amateur
Radio Club
PO Box 6000
Phoenix AZ 85005

I am in need of a schematic for a Pride 150 bilinear amplifier, 80-10 meters. If anyone has one and could send a copy, it would be a great help.

Francis Whittier WB1CX
RFD #1 Box 339
Madison ME 04950

LETTERS

from page 14

electronics experimenter, though, and guess I must agree with you that, in my experience, your mag is more with it in terms of construction articles than are the official publications of the ARRL. I tuned in too late to know the whole background of your seeming vendetta against this organization, but I suspect that you have some valid points, if only because too many organizations in and out of gov't (plus religions) get hardening of the arteries too soon, after which everyone wants to merely survive and not rock the boat. Being an iconoclast, I'm sympathetic.

Recalling the ARRL "pledge of allegiance," so to speak, that would have the ham saying that he owed his hobby to the ARRL, much as I owe my soul to the company store, I don't think I'd care to be represented by that "religious" denomination even if I were to become a ham.

Beyond that, there was a time when I used to buy the *Handbook* every few years, mainly for new test instrument circuits to build. Then I noted that it was about the last and least source of such information. Now I get the unchanging data I need (like the nice LCR resonance chart) from editions some 10 years old. The formula is still good, but magazines such as yours give me a better update on test instrument circuits, even though I scarcely need your July antenna issue, except maybe for my TV set.

Benjamin Johnson
Rochester NY

ECLECTICISM

It may interest the ham community to know that Sears, after seemingly abandoning ham radio in the Spring-Summer '79 catalog, has jumped back in with both feet.

On pp. 1238-9 of the Fall-Winter '79 catalog, the Sears 22-channel 2m FM xtal rig has returned, grouped with—brace yourself—the Midland (that's right, Midland—not a Sears nameplate in sight) 13-510A 2m syntho rig and the Midland (again!) 13-509 and 13-513 220(!) FM rigs (xtal and syntho respectively). I'm not going to comment on prices (other than to

say that they seem reasonable) because there is a bigger matter which this raises.

I was perusing *Looking West* in '73 #200 (May, 1977) which touched on the flap raised by Sears' first foray into amateur gear. This increased involvement (especially the 220 rigs) is bound to raise even more, so I thought I'd get my two cents in before inflation struck. At least Sears does point out that you must have a license, although I wish they could get it straight that it's *Technician*, not "Technical." I'm a Tech myself, and I like them to get my class right. What's more, the added exposure of 220 (use it or...) can't hurt! As for the bugaboo about "da CBer gonna getcha"—well, if we can't handle a few unauthorized signals (the price of the rigs will keep the CB-boom \$39 radio crowd off the band), we are in trouble.

On other matters, keep the mag the way it is (if not better!). Don't be bullied away from the eclecticism, the controversy, and the general free attitude of the magazine! At this time, it's the only ham magazine (counting *Worldradio* as a newspaper) worth reading! Keep it up!!

Edward Eastman WD8ENF
Minneapolis MN

IGNORAMI

I cannot sit idly by any more and let some unintelligent and uneducated hams slash your wonderful magazine and enlightening editorials. I cringe every time someone says that Wayne is against ham radio (WD8DWO, Letters, '73 Magazine, June, '79). These people must have read his editorials with their dunce caps on! Anyone who digests them with intelligence or has read Wayne's writings for a while can clearly see that Wayne Green is one of the leading proponents of ham radio and that ham radio is one of Wayne's lifetime loves.

I have been reading his editorials since I received my Novice license in September of '75, and while there have been times when I disagreed with Wayne, I have always been impressed with his honesty and his factual unbiased reports of many situations. I admire the way he has never backed down

on an issue, nor been afraid to state his opinion on any issue. I also admire the way he has succeeded in his business ventures.

So the next time you read one of Wayne's dynamic editorials, read it intelligently and you will learn about the inside happenings of ham radio, politics, business, and about devotion. I have learned about these things and more from Wayne's editorials, and so even though I know Wayne will not let a few ignorami stop him, count me as one of his supporters and keep those unabashed, interesting, and provocative editorials coming. Your magazine is the best in ham radio and better than most of the others in other fields, so continue the good work and take it easy!

John Kerekes N9EI
South Bend IN

GREEN PAGES

I just received my first '73 issue on a new subscription. The article on "The KGCY Story" was worth the price of the subscription. Keep articles like this coming and I'll renew ... could do with fewer "green" pages, however.

Ron Schwendt N3AR
Douglassville PA

Egyptian Progress

As the first woman member of the Egyptian Radio Club, Inc., Granite City IL, I feel compelled to respond to the "Letter to the Editor" written by Tania Miller WB9TKC that appeared in the August, 1978, issue of '73 Magazine. Tania's letter was written as a result of the difficulty experienced by women who wished to become members of the club. At that time there were enough members who did not want women in the club to block any attempt by women to join. Since that time, however, open-minded members have been responsible for changes that allow any responsible amateur radio operator to join the club.

I believe that this progress shows that old traditions and ideas must be changed when they do nothing except cause hurt feelings or give people in control a feeling of power. There is no place in amateur radio for prejudice.

My husband and I have both enjoyed being members of the Egyptian Radio Club very much. We have met and learned from a super bunch of fellow hams. We are pleased with the progress that has been shown in the Egyptian Radio Club, Inc.

Bess J. Nelson WD8CZF
Florissant MO

NUKES

One year ago, I was one of those people who thought nuclear power would really help. I didn't question the authority of the geniuses who gave us Three Mile Island, Brown's Ferry, and the Karen Silkwood cover-up.

Since then I have learned a lot about the poor engineering, short-term monetary greed, and media whitewashing that goes hand in hand with the nuclear industry.

On June 30, I and forty to fifty thousand people were at the Diablo Canyon rally in San Luis Obispo, California. I saw exhibits there on solar and many types of alternative energy—alternatives that have been here for years. There were people from all walks of life there, and even some fellow amateurs.

I believe this is no longer a small minority issue. I feel that we hams should set a good example and use our excellent tinkering and engineering ability to apply to solar and alternative electricity sources.

Let amateurs use solar power at the home QTH, at Field Day, and at outdoor fairs.

Let amateur publications show conscience and responsibility by running how-to articles on alternative energy. Let amateurs, through their global communications capabilities, exchange information on solar and other alternative electric power.

I hope you will take these ideas to heart for technical articles, parts source articles, and editorials. I am one ham who doesn't need "nukes," and I know there will be others.

James G. Coote WB6AAM
Los Angeles CA

CONVERTING

Perhaps this information will be of interest to those of you converting 40-channel CBs to 10 meters. I received a Hy-Gain model 2681 board, less channel switch (which includes rotary dial and knob). I found that Hy-Gain parts are available through Telex Communications, Inc., 8601 Northeast Highway 6, Lincoln NE 68505, Attn: Jim Kepustka.

The parts noted above are listed as: part #700047, switch, \$0.75; 40-ch dial, \$0.50; channel knob, \$0.50; shipping (very good box), \$1.00.

The board is sold by Poly Pak, and costs about \$12.00. They will supply a schematic diagram (specifically request it). Other info is supplied but it does not have much to do with the board received. It does give

info on another type of PLL which may or may not be used in another board they can furnish for more money. A complete Hy-Gain 40-channel set with all controls on the mike is available; the mike is an extra.

I haven't yet seen an SSB rig on sale through any of my contacts for less than \$65. One should be available soon, as the popularity of CB in the crowded cities is going down fast.

The big surprise is that the going price for a divider-type switch used with other rigs seems to be from \$16 to \$25. I get that from the CB repairmen, who say that they practically never have to replace them. Look at the price I paid for it (75 cents). When I figured out how to build a switch to divide by 278 or so, I thought \$20 was pretty fair.

Henry B. Plant W6DKZ
San Jose CA

CLASSICS

Being a Novice of about three months, I find my subscription to 73 thoroughly satisfying and enlightening in all aspects. It is very gratifying to have a publication on our behalf that pulls no punches and has its cards laid out on the table instead of putting up a smoke screen and being cloaked with the "official final authority" attitude approach. In your June issue (#225, page 13) Letters department, Don Hurley VE3HAN suggested an "equipment evaluation" booklet of rigs dating back over the past twenty years. Sounds like a great idea. Obviously an enormous amount of time and effort would need to be put into a big project on that order. Consider, though, how much value such a reference-type booklet like that could be... it would be ideal to bring along to that next hamfest for when you're rummaging

through those tables of used Hallicrafters, Hammarlunds, Johnsons, Nationals, etc. If an "equipment evaluation" booklet is not conceivable, then how about running a feature article in 73 on some of those more popular older classic rigs?

Speaking of so-called outdated gear, if you tune around the bands listening carefully, there are plenty out there sporting their well-maintained vintage equipment... just another facet of the art. I run those "oldies but goodies," a Viking Valiant I transmitter with a Hammarlund HQ-100 receiver. Keep up the superb work.

Bill Wolf KA2EEV
Newark NJ

FEARL

Any hams working in the military or for the military services as civilians should definitely bring their gear along on permanent changes of station to Japan. Write to FEARL, c/o Sam Fleming KA2SF, USAGH D10, APO SF 96343, or call 228-4703 (military number in Japan) for info. Licensing is done by US Forces, Japan. All US license classes get full privileges, except that no mobile portable operation is allowed (except MARS, where we have 2m frequencies and repeaters, etc.).

SFC W. J. Graham KA2WG
Signal Act-N
APO SF 96343

OLDIE

I would like to comment on your review concerning the Daiwa CN-720 swr and rf power meter (June, page 24). Your statement, "... why someone didn't think of it before..." left me bewildered.

On page 386 of the September, 1952, issue of *DL-QTC*—

yes, 1952—G. Blaesi DL1GX describes a standing-wave indicator which uses a cross-pointer instrument similar to the one used in the CN-720.

DL1GX's work was based on the work of a Mr. Buschbeck. Mr. Buschbeck's idea and invention was published in Volume 61, April, 1943, of *Hochfrequenztechnik und Elektroakustik*, decimal classification DK 621.717; 621.396.61.

In the June issue of the *Technical Library* of the Northwest German Radio Network, 1949, Dr. Roland Walter describes this type of instrument again because, due to the war, most of the original material got lost or was destroyed.

Telefunken marketed a version for use on high-power BC and shortwave transmitters in the early 50s, with power handling capacities of up to 100 kW. It was called "supervisory equipment for feedlines for short, medium, and long waves." It became a standard on all major transmitting equipment in DL-land.

Ergo, the idea of the cross-pointer instrument in the context of swr measurements is about 36 years old.

Kurt U. Grey VE2UG
Sept Iles, Quebec

MISS AMERICA

This is to advise of the following station activity: Special Event—Miss America Pageant, Atlantic City, New Jersey. Station K2BR will be operating from the Miss America Pageant Headquarters in Atlantic City NJ, September first through eighth, 1979. It will be sponsored by the Southern Counties Amateur Radio Association (SCARA). Approximate frequencies: CW—3560, 7060, 14060, 21060; Novice—3730, 7130, 21130; Phone—3935, 7235, 14280, 21380. QSL to K2BR, 591

White Horse Pike, Egg Harbor NJ 08215 (SASE, please). Traffic to and from the Miss America contestants will be accepted.

Henry G. Rainville K2HG
Miss America Pageant/
SCARA coordinator
Ventnor NJ

HAMMARLUND

I have been trying hard to find a replacement variable tuning capacitor for my Hammarlund HQ-170A VHF receiver and finally found one at Pax Mfg., 100 Montauk Hwy., Linden Hurst NY 11757, (516)-884-4300.

In talking with Mr. Peter Kjeldsen of Pax Mfg., he indicated that they have most replacements for Hammarlund capacitors. Maybe this information will help someone else find parts.

Leroy Marion W8CGQ
Marlette MI

JAMMING

As printed in "Letters" of the June, 1979, issue of 73, reader K6EGM is "... appalled that you could publish something ("The 2-Meter ECM Caper," 73, February, 1979) tantamount to sanctioning the jamming of another station..." and says that such "new" ideas are not in the best interest of amateur radio.

Apparently K6EGM is not well versed in the history and romance of amateur radio. "Jamming," intentional, accidental, or fantasized, is not a "new" idea. It has been going on for over 50 years in every and any call area.

I suggest that K6EGM read "The Templeton Case" by K9ODE which appeared in the January, 1963, issue of QST magazine.

Byron H. Kretzman W2JTP
Huntington NY

New Products

from page 40

The new pocket scanner features sturdy construction, with an anodized aluminum front panel to withstand demanding on-the-go use. A flexible "rubber ducky" antenna is supplied, but the radio can also be used with a wire antenna. The radio can be operated from external power, as well as from internal batteries. Also contributing to the radio's versatility are provisions for plugging in an external battery charger, headphone, and external speaker. For further information, contact *Electra*

Company, PO Box 29243, Cumberland IN 46229. Reader Service number E40.

TRAC CMOS ELECTRONIC KEYER

TRAC Electronics, Inc., has introduced an addition to its line of state-of-the-art CMOS keyers. The TRAC CMOS Electronic Keyer, Model TE133, contains all CMOS integrated circuitry. Included features are self-completing dots and dashes; both dot and dash memory; iambic keying with any squeeze paddle; 5-50 wpm; speed, vol-



TRAC's Model TE133 keyer.

ume, tune, and weight controls; and built-in sidetone and

speaker. Low-current-draw CMOS allows portable battery

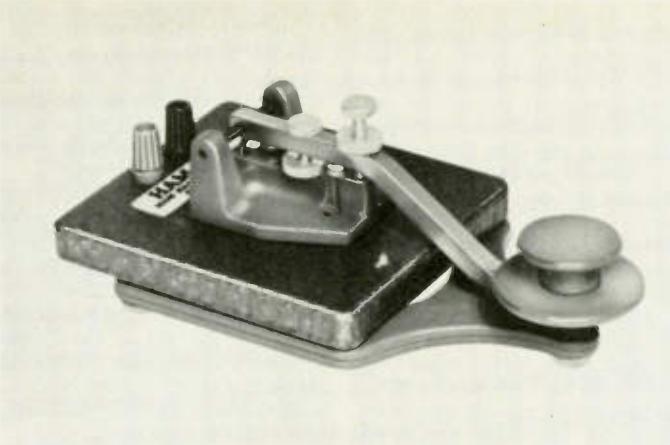


Heath/Schlumberger's new catalog.

operation. The rear panel contains deluxe quarter-inch jacks for output and keying. The TE133 CMOS Electronic Keyer is operated on a single 9-volt battery and keys both positive and negative keyed rigs. For further information, contact *TRAC Electronics, Inc., 1106 Rand Building, Buffalo NY 14203.* Reader Service number T18.

HEATH/SCHLUMBERGER OFFERS LATEST INSTRUMENTS CATALOG

Heath/Schlumberger has announced the publication of its latest instruments catalog. This new catalog features Heath/Schlumberger's complete line of fully assembled and tested computers and peripherals and gives complete descriptions



Ham Radio Center's HK-3M.

and specifications for their line of electronic test instruments. Included are oscilloscopes, laboratory-grade strip and X-Y recorders, power supplies, various signal and function generators, counters, and a full line of multimeters (from analog to digital).

In addition, the publication contains a complete listing of Heath/Schlumberger Continuing Education Programs for industrial training, including ac and dc electronics, semiconductor devices, digital techniques, microprocessors, and test instruments.

For further information, contact *Heath/Schlumberger, Dept. 350-870, Benton Harbor MI 49022.* Reader Service number H5.

THE HK-3M KEY

Ham Radio Center has introduced a new modified straight key which replaces their popular Model HK-3 Ham-Key™. The new model is the HK-3M; it features a new anti-tip bracket that defies even a pump-handle type of operator from tipping over. The beauty of this new feature is that any HK-3 now in the field can be converted to an HK-3M by merely adding the easily-installed AT-B bracket.

For further information, contact *Ham Radio Center, PO Box 28271, St. Louis MO 63132.* Reader Service number H2.

Ham Help

I would like a schematic and input/output impedance for an RME 80-10 preselector. The model is unknown, but I think it matched the RME 6900 receiver. Thanks.

**William F. Mollenhauer N2FZ
Box 3, RFD 1
Glassboro NJ 08028**

I would like to receive sche-

matics for any modifications to the Ten-Tec PM2/2A QRP rig.

**H. Goldberg VE3JBU
PO Box 913, Stn. B
Ottawa, Ontario
Canada K1P 5P9**

I would like to get in touch with amateurs who believe in UFOs, who have made a sighting or who are members of any organization that investigates UFOs. I would like to organize an international net.

**Anastasios Panos SV1IG
PO Box 2563
Athens, Greece**

I got a flood of replies to my request in the June issue. Thanks a lot, fellows! Instead of a brief visit in August, I have decided to come on the "HOUSA" student exchange/work scheme from November, 1979, to February, 1980. Under this scheme, NZ students can come to the US or Canada to work and sightsee.

Ideally, I would like to work

for short periods in various parts of the US, and would appreciate information on temporary jobs and accommodations.

I would also like to hear about UHF FM repeaters, as I intend bringing along my home-brew mobile rig (cheap, but good), based on our club design. Do UHF repeaters need the special access techniques, e.g., tone burst, etc., like your 2-meter ones?

**Ash Nallawalla ZL4TBU
PO Box 6159
Dunedin, New Zealand**

I have been following the articles on the conversion of CB sets to 10 meters and I have been trying to convert a SBE Sidebander II mobile rig, but have run into problems.

Does anyone have any information on the conversion of this unit? I did find the article on the SBE Sidebander III model (January, 1979, 73), which is strictly sideband (no AM) and am told that the boards are entirely different.

I have changed the base crystals up 1.535 MHz to put channel 1 on 28.500, but, after

complete alignment, which makes the set very hot on both receive and transmit, we find ourselves on a lower frequency than even 11 meters. Around 25.535.

**James W. Barnes WB7PKR
PO Box 283
Mesa AZ 85201**

I need a schematic or conversion info on an ARC R-19 aircraft receiver which tunes 118-148 MHz. Expenses will be reimbursed.

**Howard S. Robb AFOW
Box 17
Bird Island MN 55310**

I need a manual, or a copy, for a Gonset G-76 transceiver. I will pay costs.

**Tim Rulon WA2KQD
12 Morahopa Rd.
Centerport NY 11721**

I would like to get in touch with anyone who shares my interest in model railroading for a 40m CW sked on the Novice portion.

**Rick Todd KA8AKL
14470 Basslake Rd.
Newbury OH 44065**

Social Events

Listings in this column are provided free of charge on a space-available basis. The following information should be included in every announcement: sponsor, event, date, time, place, city, state, admission charge (if any), features, talk-in frequencies, and the name of whom to contact for further information. Announcements must be received two months prior to the month in which the event takes place.

GEORGETOWN IL SEP 1-2

The 1979 Danville, Illinois, Area Hamfest will be held on September 1-2, 1979, at the Georgetown, Illinois, fairgrounds, located ten miles south of Danville on Illinois Rt. 1. Gates open at noon on Saturday for vendors to start setting up their displays. Gates open to the general public at 6:00 am Sunday. Facilities will consist of a large enclosed building 50 x 150 feet with electrical hookups available at no charge. Please bring your own tables and chairs and power cords. Outside space is also

available at a \$2.00 per person gate charge. Overnight camping on the fairgrounds is available at \$5.00 per vehicle. For information, contact Bob Wilson K9RBW, c/o Illiana Repeater Systems, Inc., PO Box "G", Catlin IL 61817.

TEXAS CITY TX SEPT 1-2

The Tidelands Amateur Radio Society (TARS) will hold its Hamfest '79 on September 1-2, 1979, at the Nestler Civic Center on 5th Street, Texas City, Texas. The registration and hospitality period will be held on Saturday evening, with the final prize drawing to be held on Sunday afternoon. There will be many small prizes, plus TARS will be giving away the winner's choice of a Kenwood TS-120S or Drake UV-3, a Wilson Mark II HT, and a Wilson System III beam. For further information and pre-registration, write Hamfest '79, PO Box 73, Texas City TX 77590.

PENSACOLA FL SEP 2

The Five Flags Amateur

Tell them you saw their name in 73

Fontana CA

We carry the following: ICOM, Midland, Amcom, DenTron, KLM, Swan, Drake, Ten-Tec, Wilson, SST, MFJ, Hy-Gain, Lunar, Nye-Viking, B&W, Redi-kilowatt, CusCraft, Mosley, Big Signal, Pipo, etc. Full Service Store **Fontana Electronics**, 8628 Sierra Ave., Fontana CA 92335, 822-7710.

Santa Clara CA

Bay area's newest Amateur Radio store. New & used Amateur Radio sales & service. We feature Kenwood, ICOM, Wilson, Yaesu, Atlas, Ten-Tec & many more. Shaver Radio, 3550 Lochinvar Avenue, Santa Clara CA 95051, 247-4220.

Denver CO

Experimenter's paradise! Electronic and mechanical components for computer people, audio people, hams, robot builders, experimenters. Open six days a week. **Gateway Electronics Corp.**, 2839 W. 44th Ave., Denver CO 80211, 458-5444.

Columbus GA

KENWOOD—YAESU—DRAKE
The world's most fantastic amateur showroom! You gotta see it to believe it! Radio Wholesale, 2012 Auburn Avenue, Columbus GA 31906, 561-7000.

Preston ID

Ross WB7BYZ, has the Largest Stock of Amateur Gear in the Intermountain West and the Best Prices. Call me for all your ham needs. **Ross Distributing**, 78 So. State, Preston ID 83263, 852-0830.

Terre Haute IN

Your ham headquarters located in the heart of the midwest. **Hoosier Electronics, Inc.**, 43B Meadows Shopping Center, P.O. Box 2001, Terre Haute IN 47802, 238-1456.

Littleton MA

The ham store of N.E. you can rely on. Kenwood, ICOM, Wilson, Yaesu, DenTron, Whistler radar detectors, Bearcat, Regency, antennas by Larsen, Wilson, Hustler, GAM, TEL-COM Inc. Communications & Electronics, 675 Great Rd., Rt. 119, Littleton MA 01460, 486-3040.

Laurel MD

We stock Drake, Icom, Ten-Tec, Swan, Tempo, Dentron, DSI, KDK, Wilson, Midland, and others. Call toll free 800-638-4486. The Comm Center, Inc., Laurel Plaza, Rte. 198, Laurel MD 20810.

St. Louis MO

Experimenter's paradise! Electronic and mechanical components for computer people, audio people, hams, robot builders, experimenters. Open six days a week. **Gateway Electronics Corp.**, 8123-25 Page Blvd., St. Louis MO 63130, 427-6116.

Camden NJ

X-Band (& other frequencies) Microwave Components & Equipment. Laboratory Grade Test Instruments, Power Supplies, 1000's in stock at all times, BUY & SELL all popular models—HP, GR, FXR, ESI, Sorensen, Singer, etc. **Lectronic Research Labs**, 1423 Ferry Ave., Camden NJ 08104, 541-4200.

Radio Association, Inc., will hold its 1979 Ham-A-Rama on September 2, 1979, at the Pensacola Municipal Auditorium, Pensacola, Florida.

ROSEMONT IL SEP 7-9

The Quarter Century Wireless Association will hold its 1979 Chicago Convention on September 7-9, 1979, at the O'Hare/Kennedy Holiday Inn, Rosemont, Illinois. The complete package for the three days is \$35.00. Special room rates will also be available. There will be the annual banquet, special ladies' program, various tours, and prizes. For reservations and information, write Phil Haller W9HPG, 6000 S. Tripp, Chicago IL 60629.

BLOOMINGTON IN SEP 8

The second annual Hoosier Backyard Hamfest will be held on September 8, 1979, at the Hensonberg School, just east of the intersection of state highway 37 and Vernal Pike, in Bloomington, Indiana. Tickets are \$1.00 per head over age 12. Features of the event will include ATV, an ATV repeater, SSTV demonstrations, a home-computer show, inside swap area, plenty of parking, food, and some door prizes. Talk-in

on 147.78/18. Send all inquiries to HBYH, 7391 W. Hwy. 46, Ellettsville IN 47429.

AUGUSTA NJ SEP 8

The Sussex County Amateur Radio Club will hold its hamfest on Saturday, September 8, 1979, from 9:00 am to 5:00 pm, rain or shine, at Sussex County Farm and Horse Show grounds, off Rte. 206, Augusta, New Jersey. There will be a large indoor selling area and tailgating. Admission for buyers is \$1.00, which includes a chance at the door prizes. YLs, XYLs, and harmonics will be admitted free. Admission for indoor sellers is \$6.00 at the door and \$5.00 in advance. For tailgaters, admission is \$5.00 at the door and \$4.00 in advance. Talk-in on 147.90/.30 and 146.52. For registration and information, write Sussex County Amateur Radio Club, PO Box 11, Newton NJ 07860, or call Ed Woznicki AC2A at (201)-852-3268.

UNIONTOWN PA SEP 8

The Uniontown Amateur Radio Club will hold its 30th annual Pie Gabfest on September 8, 1979, starting at noon at the club grounds, Old Pittsburgh Rd., at the bypass on Rte. 51.

DEALER DIRECTORY

Tell them you saw their name in 73

North Arlington NJ

Terre Haute IN

Collins, General Radio, Tektronix, Hewlett Packard. Radio sales/service. Cash or trade for tubes or gear. **Fed W2KUW, DCO, INC.**, 10 Schuyler Avenue, No. Arlington NJ 07022 998-4246.

Syracuse—Central NY

We deal, we trade, we discount, we please! Yaesu, Ten-Tec, CusCraft, Drake, Dentron, KLM, Midland, B&W, ICOM, Hygain, Swan, Amcom, Telco, Mirage, DSI etc. Complete 2-way service shop! **Ham-Bone Radio (Div. Stereo Repair Shop)** 3206 Erie Blvd. East, Syracuse NY 13214, 446-2266.

Syracuse-Rome-Utica NY

Featuring: Yaesu, ICOM, Drake, Atlas, DenTron, Ten-Tec, Swan, Tempo, KLM, HyGain, Mosley, Wilson, Larsen, Midland Southwest Technical Products. You won't be disappointed with equipment/service. **Radio World, Oneida County Airport-Terminal Building**, Oriskany NY 13424, 337-2622.

Scranton PA

ICOM, Bird, CusCraft, VHF Engineering, Antenna Specialists, Barker & Williamson, CDE Rotators, Ham-Keys, Belden, W2AU/W2VS, Shure, Regency, CES Touch-Tone pads, Radio Amateur Calbooks. **LaRue Electronics**, 1112 Grandview St., Scranton PA 18509, 343-2124.

Houston TX

Experimenter's paradise! Electronic and mechanical components for computer people, audio people, hams, robot builders, experimenters. Open six days a week. **Gateway Electronics Inc.**, 8932 Clarkcrest, Houston TX 77063, 978-6575.

Port Angeles WA

Mobile RFI shielding for elimination of ignition and alternator noises. Bonding straps. Components for "do-it-yourself" projects. Plenty of free advice. **Estes Engineering**, 930 Marine Drive, Port Angeles WA 98362, 457-0904.

DEALERS

Your company name and message can contain up to 25 words for as little as \$150 yearly (prepaid), or \$15 per month (prepaid quarterly). No mention of mail-order business or area code permitted. Directory text and payment must reach us 45 days in advance of publication. For example, advertising for the November issue must be in our hands by September 15th. Mail to 73 Magazine, Peterborough NH 03458. ATTN: Aline Couto.

Uniontown, Pennsylvania. Registration is \$2.00, with a chance for a main prize of a Tempo S-1. There will also be a Swap 'n Shop.

MENA AR SEP 8-9

The QWHC is celebrating the 10th anniversary of the Queen Wilhelmina Hamfest on September 8 and 9, 1979, at the Queen Wilhelmina State Park lodge, atop Rich Mountain just north of Mena, Arkansas. Featured will be displays, a free flea market, a banquet, a live band, church services, activities for XYLs and harmonics, and much, much more. There will be a grand prize of a Yaesu FT-101ZD, and many more prizes. Look for WB5MFI/P5 operating from the lodge. Admission is \$2.00. Tickets for the grand prize are \$2.00 each or 3 for \$5.00. Talk-in on 3995, 146.19/.79, and 146.52. For additional information, contact Steve Myers WB5MFI, SS103 Carlson Terrace, Fayetteville AR 72701, or phone (501)-443-3489.

MONTGOMERY AL SEP 8-9

The 2nd annual Central Alabama Hamfest is scheduled for September 8-9, 1979, at the

Montgomery Civic Center, 300 Bibb St., Montgomery, Alabama. The FCC will administer exams beginning at 8:00 am on Saturday, September 8. Bring your Form 610. All activities will be indoors in air-conditioned comfort. There will be prizes and a ladies' program. Talk-in on 146.04/146.64. For more information, contact Ed Sensin-taffar WA4NKU, 745 Dubuque Dr., Montgomery AL 36109 or Sam Windham WB4RGX, 1834 Shoreham Dr., Montgomery AL 36106.

FINDLAY OH SEP 9

The Findlay Radio Club will hold its 37th annual Findlay Hamfest on Sunday, September 9, 1979, at Riverside Park, Findlay, Ohio. There will be both commercial and amateur display space available. Ticket donation is \$1.50 in advance and \$2.00 at the hamfest site. For more information, write the Findlay Radio Club, c/o Randy Peterson, Hamfest Chairman, 6016 Marion Twp. 243, Findlay OH 45840.

BUTLER PA SEP 9

The Butler County Amateur Radio Association, Inc., will

TELREX "Tri-Band" — \$415.00

ELECTRICAL

	TB5EM
Gain in DB reference 1/2 wave dipole	8.5 dbd
Front to back ratio	28 db
V/S/W/R at resonant point	1.3/1
Maximum Power Input	4 KWP
Nominal Input Impedance	52 ohm
Beamwidth to 1/2 power Input	60° 3 bands
Frequency range	10. 15. 20
Side Nulls	35 db

MECHANICAL

Number of Elements	Five
Alum. Boom: Dia. & Lgth. approx.	2.2.5" x 18 ft.
Turning Radius approx.	20 ft.
Wind Load at 100 mph (approx.)	210 lbs.
Wind Area	7 sq. ft.
Longest Element	36 ft.
Net Weight (approx.)	49 lbs.
Shipping weight (domestic pack)	60 lbs.
Length of shipping carton	13 ft.

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hold "Ye Olde Fashioned Hamfest" on September 9, 1979, from 10:00 am to 4:00 pm at the Butler County Farm Show Grounds adjacent to Roe Airport on Rte. 68, west of Butler, Pennsylvania. The \$1.00 charge includes free parking and admission to the outside flea market. Children under 12 are admitted free. Overnight campers are welcome. Tables for the indoor flea market are \$3.00 for an 8-foot table provided by us, or \$2.00 for an 8-foot table provided by you. Food and refreshments will be available. Mobile check-ins on .52 and 147.90/.30 (W3UDX). Fly-ins welcome (80 + 100 aviation gas will be available). YL, mobile, and fly-in prizes will be awarded. There will be six main prizes plus other prizes to be drawn every 15 minutes. For more information, contact Fred Young WB3HGC, 195 Robbie Way, Portersville PA 16051, or phone (412)-368-3386.

PECATONICA IL SEP 9

The Rockford Amateur Radio Association will hold its second annual Rockford Hamfest and Illinois State ARRL Convention on Sunday, September 9, 1979, at the exhibition hall at the Winnebago County Fairgrounds at Pecatonica, Illinois, just west of Rockford on US Rte. 20. Tickets are \$2.00 in advance or \$2.50 at the gate. Tickets are available by mail by writing RARA, PO Box 1744, Rockford, Illinois 61110. Please include an SASE for tickets by mail. Prizes include a Kenwood TS-520S transceiver and an Atlas receiver. Campsites are available on site, with electric and sanitary hookup available. There are 300 flea-market tables available at a nominal charge. Plenty of free parking is available. Featured will be speakers, forums, demonstrations, and discussions. A hamfest menu, including hot dogs, BBQ, and soft drinks will be available at reasonable prices. Talk-in on 146.01/.61 or 146.52.

PENNSAUKEN NJ SEP 9

The South Jersey Radio Association will hold its hamfest on September 9, 1979, at Pennsauken Senior High School, Hylton Rd. at Rtes. 3, 130, and 70. Admission is \$2.00, tailgating is \$3.00, and an inside table is \$5.00. For further information, contact Bruce Eichmann WA2NBM, Hamfest Chairman, 204 E. Lake Blvd., Marlton NJ 08053, or phone (609)-983-0106.

VALPARAISO IN SEP 9

The Porter County Amateur

Radio Club will hold its annual hamfest on September 9, 1979, at the Porter County Fairgrounds, Valparaiso, Indiana. Featured will be a flea market, prizes, and technical sessions. Admission is \$2.00. There will be no charge for a flea market space but do bring your own tables. Talk-in on 147.96/.36 and 146.52. For advance tickets and information, write Art Cushman N9FB, 944 N. 100 W., Valparaiso IN 46383.

HUDSONVILLE MI SEP 15

The Grand Rapids Amateur Radio Association, Inc., will hold its annual Swap 'n Shop on Saturday, September 15, 1979, at the Hudsonville Fairgrounds, Hudsonville, Michigan. Gates open at 6:00 am; sales begin at 8:00 am. Talk-in on .16/.76, .63/.03, and .52/.52.

PEORIA IL SEP 15-16

The 21st annual Peoria Superfest 1979 will be held on September 15-16, 1979, at Exposition Gardens, W. Northmoor Rd., Peoria, Illinois. Advance tickets are \$2.00. Door tickets are \$3.00. Camping will be available Friday night on the grounds. There will be free indoor and outdoor flea market space available. Tickets will be sold on the grounds for hourly prize drawings. Main prize drawing will be at 3:00 pm both days. Continuous demonstrations of the latest equipment and technology will be given by major manufacturers and dealers. All dealers, distributors, and manufacturers write or call for info on free commercial table space. There will be movies, forums, talks, and displays of interest to all given daily. For the ladies: There will be a free bus trip to the Northwoods Mall on Sunday at 12:00 noon, plus daily displays and demonstrations and also a ladies' flea market. An informal get-together will be held at the Heritage House smorgasbord, 8209 N. Mt. Hawley Rd., (Rte. 88) at 7:30 pm Saturday, September 15th. No reservations are necessary. Price will be approximately \$5.75 per person. Talk-in on 146.76 W9UVI and .76, .85, and .97. For advance tickets, write Peoria Hamfest, 5808 N. Andover Ct., Peoria IL 61614, or phone (309)-692-8763.

ROSS OH SEP 16

The Greater Cincinnati Amateur Radio Association, Inc., will hold its 43rd annual Cincinnati Hamfest on Sunday, September 16, 1979, at Strickers Grove on Ohio Rte. 128, one mile west of Ross (Venice), Ohio. Exhibits, prizes, food, and refreshments will be available.

Featured will be a flea market with radio-related products only, music and good fellowship, a hidden transmitter hunt, and a sensational air show. Admission and registration are \$4.00. For further information, write Lillian Abbott K8CKI, 1424 Main St., Cincinnati OH 45210.

MT. CLEMENS MI SEP 16

The L'Anse Creuse Amateur Radio Club will hold its 7th annual Swap and Shop on September 16, 1979, at the L'Anse Creuse High School, Mt. Clemens, Michigan. Take I-94 eastbound to the Metro Parkway exit. Then take the Metro Parkway to Crocker. Go left on Crocker to Reimold and then right on Reimold to the last school, L'Anse Creuse High School. Admission is \$2.00 at the door, or \$1.00 in advance. There will be plenty of food and parking plus hourly prize drawings. Prizes include a first prize of \$200.00, a second prize of \$100.00, and a third prize of \$50.00. Talk-in on 147.69/.09 and 146.52. For more information, send an SASE to WD8ITS, 3488 Ashley, Pontiac MI 48055.

HARRISBURG PA SEP 16

The Central Pennsylvania Repeater Association, Inc., will hold its 6th annual High Rise Hamfest on Sunday, September 16, 1979, from 8:00 am until 3:00 pm at the Park and Shop Garage, 200 block of Walnut Street, Harrisburg, Pennsylvania. Admission is \$3.00, with tailgating, wives, and kids free of charge. There will also be door prizes.

QUEENS NY SEP 16

The Hall of Science Amateur Radio Club, Inc., will hold its 2nd annual indoor/outdoor, rain or shine electronics hamfest on September 16, 1979, from 9:00 am to 4:00 pm at the municipal parking lot, one block off Queens Blvd. at 80-25 126th St., Queens, New York. There will be free parking, refreshments, and free prizes. Admission is \$2.00 for sellers and \$1.00 for buyers. Talk-in on .52/.52 and .96/.36.

WHITESTONE NY SEP 20

The Tu-Boro Radio Club will hold its auction on Thursday, September 20, 1979, from 6:00 pm to 10:00 pm at the Odd Fellows Hall, 149-14 14th Avenue, Whitestone, New York. Donation is \$1.00. Talk-in on 145.62 and 146.52.

ELMIRA NY SEP 22

The Elmira Amateur Radio Association will hold its fourth

annual International Hamfest on September 22, 1979, from 8:00 am to 5:00 pm at the Chemung County Fairgrounds, Elmira, New York. There will be prizes, programs, a free flea market, and dealers' and manufacturers' displays. Talk-in on .52, .96/.36, and .10/.70. For more information, contact John Breese WA2FJM, 340 West Avenue, Horseheads NY 14845.

ADRIAN MI SEP 23

The Adrian Amateur Radio Club will hold its 7th annual hamfest on Sunday, September 23, 1979, at Lenawee County Fairgrounds, Adrian, Michigan. Featured will be prizes, games, and programs. Tables are available at \$5.00 per 8 ft. space, \$3.00 per 4 ft. space, \$1.00 per 8 ft. trunk space, and \$2.00 for an inside space. Talk-in on 146.31/.91 and 146.52. For ticket and table information, write Adrian Amateur Radio Club, Inc., PO Box 26, Adrian MI 49221, or call Bob or Sally Fay of Sword Enterprises at (517)-263-3597.

FLINT MI SEP 23

The Genesee County Radio Club, along with the Bay Area Radio Club, the Lapeer County

Radio Club, the Saginaw Valley, and the Shiawassee County Radio Club, will hold their 2nd annual 5-County Swap 'n Shop on September 23, 1979, from 7:30 am to 4:00 pm at Southwestern High School, Flint, Michigan. Take I-69 to the Hammerberg Rd. exit, turn south to 12th St., then left on 12th to the high school. Admission is \$2.00 for adults, with children under 12 free. There will be refreshments, prizes, and fun for all. Talk-in on 146.52, 147.27, and 146.91. For information, contact Don Williams WD8QPM, 5114 Knapp Dr., Flint MI 48506.

BEREA OH SEP 23

The fourth annual Cleveland Hamfest will be held on Sunday, September 23, 1979, at the Cuyahoga County Fairgrounds, Berea, Ohio. The hamfest will be an all-indoor operation. There will be 10-foot booths available with an 8-foot table and two chairs for \$30.00.

GAINESVILLE GA SEP 23

The Lanierland Amateur Radio Club will hold its sixth annual Hamnic at the Lake Lanier Islands Dogwood Pavilion on September 23, 1979. There will be two large covered pavilions

and a large parking area for the swap shop and exhibits. Food will be available. There will be no entry fee for Hamnic; however, Lanier Islands charges a \$2.00 entry fee per car. There will be picnicking, hiking, and swimming available for the kids. Trailer hookups and camping are also available on site. First prize will be a KDK FM2015R; there will be many other prizes. Talk-in on W4IKR .07/.67. For further information, write Bob Cochran W4DNX, 607 East Lake Drive, Gainesville GA 30501.

BOULDER CO SEP 23

The Boulder Amateur Radio Club will hold Barcfest '79, on September 23, 1979, beginning at 9:00 am at the Boulder National Guard Armory, North Broadway, at the city limits, Boulder, Colorado. There will be an auction and a snack bar. Admission is \$2.00 which includes a door prize drawing. Talk-in on 146.10/.70 and .52/.52. For further information, contact Mark Call N0MC, 4297 Redwood Ct., Boulder CO 80301, or phone (303)-442-2616.

SUTTON NH SEP 23

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103A	111	175	225
104	106	176	226
105	2.27	177	228
106	80	179	229
107	79	180	230
108	89	181	231
121	2.15	182	232
123	69	183	233
123A	79	184	234
124	1.53	185	235
126	1.16	186A	236
127	4.80	187A	237
128	1.37	188	238
129	1.56	189	239
130	1.95	190	241
131	1.98	191	247
132	0.71	192	276
133	1.14	193	1.04
152	1.43	194	82
153	1.85	195A	2.67
154	1.85	196	1.98
155	2.02	197	1.89
157	1.43	198	1.89
158	1.08	199	59
159	86	210	1.37
160	1.43	211	1.56
161	98	218	3.08
162	5.75	219	4.36

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Association will hold its hamfest on September 23, 1979, from 9:00 to 5:00 pm at King Ridge ski area, Exit 11 off I-89, Sutton, New Hampshire. Admission at the gate is \$2.00 per person over 16. Advance admission is \$1.50. Send to C. A. Breuning, 54 Myrtle Street, Newport NH 03773. Festivities include an indoor/outdoor flea market, a florist exhibit, a frisbee toss, a horseshoe competition, dealers' exhibits, food, overnight camping available for self-contained units only, and a consignment room. Door prizes will be awarded, including a grand prize of a handheld, portable synthesized 2-meter rig, plus a raffle of a 19" color television at 5:00 pm. Talk-in on .52/.52 and .16/.76.

ERIE PA SEP 23

The Radio Association of Erie, Inc., will hold its annual Ham Jam '79 on September 23, 1979, between the hours of 8:00 am and 4:00 pm at Rainbow Gardens, Waldemeer Beach Park, at the head of Presque Isle Bay, Erie, Pennsylvania. Tables can be reserved on a first-come, first-served basis, at a cost of \$5.00 per table. There will be plenty of free parking and lots of indoor display area. Admission is \$3.00 at the gate and includes a door-prize ticket, a YL ticket, a main-prize ticket, and a free coffee. Commercial displays will also be available. There will be a \$1.00 charge per car for display space in the flea market area. Refreshments will be served on the premises. Talk-in on 146.34/.94, 146.34 simplex, and 7,250 kHz. For additional information or advanced tickets, send an SASE to Ham Jam '79, c/o the Radio Association of Erie, PO Box 844, Erie PA 16512.

LOUISVILLE KY SEPT 29-30

The ninth annual Greater Louisville Hamfest and Kentucky State ARRL Convention will be held on September 29-30, 1979, at the West Hall of the Kentucky Fair and Exposition Center in Louisville, Kentucky. Follow QSY signs off either I-65 or I-264. There will be a gigantic indoor air-conditioned exhibitors' area and flea market plus meetings, forums, and ladies' program. FCC exams will be given on Saturday with a banquet on Saturday night. Admission is \$3.00 in advance, \$3.50 at the door, and \$9.75 for the banquet. Exhibitors and flea market vendors should write for a special information sheet. For other information, contact the Greater Louisville Hamfest, PO Box 34444, Louisville KY 40232, or phone (502)-634-0619.

WARNER ROBINS GA SEP 30

Central Georgia ARC's first annual hamfest will be held on September 30, 1979, from 8:00 am to 5:00 pm at the City Recreation Center, on Watson Boulevard, Warner Robins, Georgia. Dealers and flea market will be indoors. The Georgia Single Side Band Association and Georgia CW Association both plan their annual meetings with us. Talk-in on 3,975 LSB and 146.25/.85 and 146.52. For more information, call or write Bill Atkins WD4ASB, 201 Avalon Drive, Warner Robins GA 31093, or phone (912)-923-3454.

GAITHERSBURG MD SEP 30

The Foundation for Amateur Radio will hold its annual hamfest at the Gaithersburg Fairgrounds, Gaithersburg MD, on Sunday, September 30, 1979, rain or shine. Featured are a large flea market, food service, exhibits, ladies' events, a supervised children's program, and many prizes. The main events are all indoors. Picnic grounds and free parking are available. Participation fee is \$2.00, sales space for flea market is \$6.00, and for tailgate, \$5.00 (spaces available on a first-come basis). Commercial exhibitors' spaces are \$15.00 with pre-registration required. Talk-in service provided and nearby motel rooms available. For information, write or call Ron Levin W3GBU, 802 Greenview Court, Reisterstown MD 21136, or telephone (301) 833-1816.

BLACKSBURG VA OCT 1-6

Two expanded workshops on 8080/8085/Z80 microcomputer design, microcomputer interfacing, software design, and digital electronics are being given by the editors of the popular Blacksburg books. Participants have the option of retaining the equipment used in these courses. Dates are October 1-6, 1979. For more information, contact Dr. Linda Lefel, C.E.C., VPI and SU, Blacksburg VA 24061 or phone (703) 961-5241).

SIOUX FALLS SD OCT 5-7

The '79 ARRL Dakota Division Convention will be held from October 5-7, 1979, at the Sioux Falls Airport Ramada Inn, located off Exit 81 on I-29, Sioux Falls, South Dakota. Featured will be technical and operating forums, a ladies' program, an ARRL forum, a large exhibit area, and a banquet. Prizes include an advance-registration prize of a DenTron GLA-1000 amplifier, a grand prize of a Kenwood TS-820S and a second

grand prize of a Wilson System One™ antenna and WR-500 rotor. Registration is \$15.00 (\$16.00 after September 1) or \$6.00 for the convention only (\$7.00 after September 1). Talk-in on 146.16/.76. For further information and convention-rate hotel accommodations, write Sioux Falls Amateur Radio Club, Box 91, Sioux Falls SD 57101.

HOUSTON TX OCT 5-7

The Houston Area Amateurs will host the ARRL West Gulf Division Convention on October 5-7, 1979, in Houston, Texas. For further information, contact Houston Ham Conventions, Inc., PO Box 79252, Houston TX 77024, or phone (713)-466-0518 or (713)-223-3161.

WARRINGTON PA OCT 6

The Mt. Airy VHF Radio Club Inc., will hold its Hamarama '79 and Mid-Atlantic States VHF Conference on Saturday and Sunday, October 6-7, 1979. The conference will be held on Saturday, October 6, from 9:00 am to 5:00 pm at the Warrington Motor Lodge, Rte. 611, Warrington, Pennsylvania. Featured will be an all-day VHF program, a cocktail hour and get together, and a buffet dinner. Registration is \$3.00 in advance, or \$4.00 at the door, which includes the flea market. The buffet dinner is \$9.00. The flea market will be held on Sunday, October 7, from 8:00 am to 4:00 pm, rain or shine, at the Bucks County Drive-In Theatre, also on Rte. 611. Registration is \$2.00 with tailgating \$2.00 per space (bring your own table). Featured will be amateur radio equipment, electronic parts, surplus, and door prizes. Talk-in on 146.52 W3CCX. For information, write Ron Whitsee WA3AXV, Chairman, PO Box 353, Southampton PA 18966, or phone (215)-355-5730.

CORNWALL NY OCT 6

The Orange County Amateur Radio Club will hold its annual auction on Saturday, October 6, 1979, at Munger Cottage, Cornwall, New York. Admission is \$1.00 and includes a chance on a door prize. The auction begins at 1:00 pm and sellers should arrive at noon. Talk-in on .52. For further information, contact Bill Lazzaro N2CF, 11 Jefferson St., Highland Mills NY 10930.

ROME GA OCT 7

The Northwest Georgia Amateur Radio Club will hold its annual Rome Hamfest on October 7, 1979, at the Coosa Valley Fairgrounds, Rome, Georgia. Gates will open at 9:00 am.

Talk-in on 146.34/.94 and 146.085/.685. For further information, contact WB4AEG, Box 274, Adairsville GA 30103.

ROCK HILL SC OCT 7

The York County Amateur Radio Society will hold its 28th annual hamfest on Sunday, October 7, 1979, starting at 8:00 am, at Joslin Park, Rock Hill, South Carolina. Registration is \$2.75 each, or 2 for \$5.00 in advance, or \$3.00 at the gate. The main prize is a Yaesu FT-901DM. A barbecue dinner is available at the park. Talk-in on 146.43/147.03 and 146.52. For more information, write York County Amateur Radio Society, Inc., PO Box 4141 CRS, Rock Hill SC 29730.

BERIEN SPRINGS MI OCT 7

The Blossomland Amateur Radio Association will hold its fall Swap Shop on Sunday, October 7, 1979, at the Berrien County Youth Fairgrounds, north of Berrien Springs, Michigan, on US 31, beginning at 8:00 am. There will be commercial exhibits, prizes, refreshments, plenty of free parking, and display space. Space for self-contained campers, at \$3.50 including electricity, is on the grounds. Talk-in on 146.22/.82. Advance tickets are \$1.50; \$2.00 at the gate. Eight-foot tables are \$2.00 and are restricted to electronic items. For advance tickets and information, write Charles White, 1940 Union Ave., Benton Harbor MI 49022.

TAYLOR MI OCT 7

The 3rd annual Radio and Electronic Equipment Swap & Shop will be held on Sunday, October 7, 1979, from 9:00 am until 3:00 pm, at Kennedy High School, Taylor, Michigan. Admission is \$2.00. Featured will be door prizes and food. Talk-in on .93/.33, .52/.52, and .99/.39. For information, write RADAR, Inc., PO Box 1023, Southgate MI 48195.

OTTAWA ONT CAN OCT 12-14

The Radio Society of Ontario will hold its 11th annual convention at the Skyline Hotel, Ottawa, Ontario, Canada. On Friday evening, there will be a buffet and dance. On Saturday, there will be demonstrations, forums, technical sessions, a women's program, and a banquet and dance. On Sunday, there will be a flea market and delegates' meeting. For information, write PO Box 5076, Station F, Ottawa, Ontario, CAN K2C 3H3.

ASHEVILLE NC OCT 13

The Western Carolina Ama-

teur Radio Society will hold its Asheville Autumnfest on Saturday, October 13, 1979, at the Asheville Civic Center, Asheville, North Carolina. There will be ample space for manufacturers, dealers, and the flea market, which will be in another part of the arena. A concession stand will be operated by the Civic Center. All manufacturers and dealers will have separate booths. And it will be possible to drive directly to your booth for unloading.

SYRACUSE NY OCT 13

The Radio Amateurs of Greater Syracuse will hold its annual hamfest on October 13, 1979, from 9:00 am until 6:00 pm at the New York State Fairgrounds, located adjacent to I-690, 3 miles southeast of the New York State Thruway, Exit 39, one mile northwest of Syracuse, New York. For commercial exhibitors, a fee of \$15.00 will include a booth with a display counter ten to fifteen feet in length or a table and two chairs. Included in the \$15.00 fee will be two tickets to the hamfest. Accommodations are available at nearby motels or travel trailer and motor home space will be available on the grounds. Commercial exhibitors will be able to set up their displays Friday night from 7:30 to 10:00 pm or on Saturday morning from 7:30 to 9:00 am. For more information, contact Bod Edgett or Paul Dunn, exhibitor chairmen, c/o Radio Amateurs of Greater Syracuse, PO Box 88, Liverpool NY 13088.

MEMPHIS TN OCT 13-14

The Mid-South Amateur Radio Association and participating Memphis-area clubs will sponsor the Memphis Hamfest and Tennessee State ARRL Convention on October 13-14, 1979, at the Youth Building at the Mid-South Fairgrounds, Memphis, Tennessee. Featured will be forums, exhibits, a giant flea market, FCC exams, a hospitality party, and commercial and manufacturer exhibits. The display area will be open from 9:00 am to 4:00 pm on Saturday, and from 9:00 am to 2:30 pm on Sunday. Fifty trailer hookups are on the premises, which the Memphis Park Commission will rent for \$5.00 per night.

BEAVER OK OCT 14

The Beaver Hamfest will be held on October 14, 1979, at the Fairgrounds Building in Beaver OK. Doors open at 8:00 am, with registration at 10:00 am. Tickets are \$2.50 each. There will be a covered-dish luncheon, a short program at 1:30 pm, swap tables, and door prizes. Camper

hookups are nearby and the event is airport-close. Talk-in on .01/.61 and .52. For details, contact Stella Shaw WB5VUN, Box 310, Beaver OK 73932, (405)-625-3368.

LIMA OH OCT 14

The Northwest Ohio Amateur Radio Club will hold its annual hamfest on October 14, 1979, at the Allen County Fairgrounds, Lima, Ohio. Two large heated buildings will house the hamfest where tables will be available for \$3.00 each. A flea market will be held outside for free. Advance tickets are \$2.00 each. For information, send an SASE to NOARC, PO Box 211, Lima OH 45802.

WEST GHENT NY OCT 14

The Northeastern States 160-Meter Amateur Radio Association will hold its annual election and banquet on Sunday, October 14, 1979, at Kozel's Restaurant, Rte. 9H, West Ghent, New York. There will be a flea market in the rear parking lot at 1:00 pm and a roast beef dinner at 5:00 pm. All hams and XYLs are welcome. For reservations and details, contact William Derby WA5IOD, Secretary/Treasurer, 14 Plain St., Medfield MA 02052.

ISLIP LI NY OCT 14

The Long Island Mobile Amateur Radio Club, Inc., will hold its Hamfair '79 on Sunday, October 14, 1979, from 9:00 am until 4:00 pm at the Islip Speedway, Rte. 111 (Islip Ave.), one block south of Southern State Pkwy., Exit 43, or come south from the Long Island Expressway, Exit 56, Islip, Long Island, New York. There will be free parking, door prizes, and several contests. Admission is \$1.50 (non-hams are free) and \$3.00 per seller's space, which permits one person to enter. For information, call Hank Wener WB2ALW, nights, at (516)-484-4322, or Sid Grossman N2AOI, nights, at (516)-681-2194.

ANAHEIM CA OCT 19-21

The ARRL Southwestern Division Convention will be held on October 19-21, 1979, at the Sheraton-Anaheim Hotel, located at Ball Rd. and I-5, Anaheim, California. The convention will begin on Friday evening with registration and exhibits from 4:00 pm until 9:00 pm. On Saturday, registration will begin at 8:00 am and exhibits and technical sessions will run from 9:00 am until 3:30 pm. FCC testing will continue until 3:30 pm also. The ARRL Forum will be held from 4:00 pm until 5:30 pm, with a no-host

cocktail party being held until the 7:30 pm banquet. The Wouff Hong pageant will be held at 00:01 am PST on Sunday morning. At 9:00 am Sunday morning the various breakfasts will be held and the exhibits will again be open until noon. The pre-registration deadline is September 15, 1979. Advanced registration price, which includes complete program, banquet, exhibits, and technical sessions, is \$17.00, and \$19.00 at the door. The charge for the banquet only is \$12.00, and for exhibits and technical sessions, the charge is \$5.00, pre-registration; \$6.00 at the door. The ladies' program and luncheon is \$6.00, pre-registration only. For more information and pre-registration, contact Hamcon, PO Box 1227, Placentia CA, or phone (714)-993-7140.

CEDAR RAPIDS IA OCT 19-21

The 1979 ARRL Midwest Division Convention and CVARC Hamfest will be held on October 19-21, 1979, at the Five Seasons Center, Cedar Rapids, Iowa. Tickets are \$4.00 in advance or \$5.00 at the door. Forums will include FCC, ARRL, DX, antenna, AMSAT/OSCAR, FM and repeaters, microprocessors, modern CW, and more. A flea market will be held at \$5.00 per table with 150 tables available. Reservations are good for Saturday and Sunday and must be paid in advance. Pre-registrations will be taken through October 1, 1979. Setup begins at 6:00 am Saturday. FCC exams also will be given on Saturday. (Send Form 610 and copy of license two weeks in advance.) There will be many prizes, including a grand prize of a deluxe HF transceiver, a TH6DXX antenna, a HAM III rotor, and a 60-ft Rohn 25G tower. There will be a Saturday-evening banquet, with Senator Barry Goldwater K7UGA as guest speaker. There are many hotels and motels available. Talk-in on 146.34/.94. For information, write Convention, Cedar Valley Amateur Radio Club, Box 994, Cedar Rapids IA 52406.

BILOXI MS OCT 20-21

The Gulf Coast Ham/Swap Fest will be held on Saturday and Sunday, October 20 and 21, 1979, at the International Plaza, located at the west end of the Biloxi-Ocean Springs bridge on Highway 90 in Biloxi MS. Tables are \$3 per day or \$5 per weekend. Talk-in on 146.13/73 and 146.52. For information, advance tickets, and tables, contact Al Williams WD5GNR, 311½ DeMontluzin Ave., Bay St. Louis MS 39520.

READING MA OCT 20

The Quannapowitt Radio Association will hold its annual auction on October 20, 1979, at the Knights of Columbus Hall in Reading, Massachusetts. Doors will open at 10:00 am and the auction will start at 11:00 am. Food and refreshments will be available. Talk-in on 146.52. For information, call Bob Reiser AA1M at (617)-272-6219.

NORFOLK VA OCT 20-21

The fourth annual Tidewater Hamfest-Computer Show-Flea Market will be held on October 20-21, 1979, starting at 9:00 am at the Norfolk, Virginia, Cultural and Convention Center SCOPE, Norfolk, Virginia. There will be 60,000 square feet of air-conditioned exhibit and flea market tailgating space available. Featured will be ARRL meetings, DX and traffic forums, and a CW contest. FCC exams are planned for amateur upgrading on Saturday from 9:00 to 12:00 am. A special feature will be a dinner cruise and banquet on the *Spirit of Norfolk* cruise ship on Saturday night for \$16 per person, or \$30 per couple. Advance registrations are \$2.50 (include an SASE) or \$3.50 at the door. Flea market tailgate spaces are \$3.00 per day. For tickets and information, write TRC, PO Box 7101, Portsmouth VA 23707.

LONDON ONT CAN OCT 28

The London Amateur Radio Club will hold its 2nd annual Swap and Shop on October 28, 1979, from 8:00 am until 4:00 pm at Lord Dorchester High School in Dorchester, just off 401. Admission and tables are both \$2.00. Featured will be displays and prizes. Talk-in on .78/.18. For more information, write VE3CSK, RR #1, Ailsa Craig, Ontario, Canada N0M 1A0.

FRAMINGHAM MA NOV 11

The Framingham Area Radio Association will hold its indoor electronic flea market on Sunday, November 11, 1979, from 10:00 am until 2:00 pm at the Framingham Police drill shed behind the police station, Framingham, Massachusetts. From Rte. 9, take Rte. 126 south to the center of Framingham. Sellers' setup time is from 9:00 am to 10:00 am. Advance table reservations will be \$5.00, with tables available at the door for \$7.50. Refreshments will be served outside the flea market area. Talk-in on .75/.15 and .52. For information or reservations, write Framingham Area Radio Association, PO Box 3005, Framingham MA 01701.

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ELECTRONICS



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Super Basic on Cassette \$40.00

Tiny Basic Source now available \$19.00
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Coming Soon: High resolution alpha/numerics with color graphics expandable up to 256 x 192 resolution for less than \$100. Economical versions for other popular 1802 systems also.
16K Dynamic RAM board expandable to 32K for less than \$150.

RCA Cosmac Super Elf Computer \$106.95

Compare features before you decide to buy any other computer. There is no other computer on the market today that has all the desirable benefits of the Super Elf for so little money. The Super Elf is a small single board computer that does many big things. It is an excellent computer for training and for learning programming with its machine language and yet it is easily expanded with additional memory, Full Basic, ASCII Keyboards, video character generation, etc.

Before you buy another small computer, see if it includes the following features: ROM monitor; State and Mode displays; Single step; Optional address displays; Power Supply; Audio Amplifier and Speaker; Fully socketed for all IC's; Real cost of warranty repairs; Full documentation.

The Super Elf includes a ROM monitor for program loading, editing and execution with SINGLE STEP for program debugging which is not included in others at the same price. With SINGLE STEP you can see the microprocessor chip operating with the unique Quest address and data bus displays before, during and after executing instructions. Also, CPU mode and instruction cycle are decoded and displayed on 8 LEO indicators.

An RCA 1861 video graphics chip allows you to connect to your own TV with an inexpensive video modulator to do graphics and games. There is a speaker system included for writing your own music or using many music programs already written. The speaker amplifier may also be used to drive relays for control purposes.

Super Expansion Board with Cassette Interface \$89.95

This is truly an astounding value! This board has been designed to allow you to decide how you want it optioned. The Super Expansion Board comes with 4K of low power RAM fully addressable anywhere in 64K with built-in memory protect and a cassette interface. Provisions have been made for all other options on the same board and it fits neatly into the hardwood cabinet alongside the Super Elf. The board includes slots for up to 6K of EPROM (2708, 2758, 2716 or TI 2716) and is fully socketed. EPROM can be used for the monitor and Tiny Basic or other purposes.

A K Super ROM Monitor \$19.95 is available as an on board option in 2708 EPROM which has been preprogrammed with a program loader/editor and error checking multi file cassette read/write software. (relocatable cassette file) another exclusive from Quest. It includes register save and readout, block move capability and video graphics driver with blinking cursor. Break points can be used with the register save feature to isolate program bugs quickly, then follow with single step. The Super Monitor is written with subroutines allowing users to take advantage of

monitor functions simply by calling them up. Improvements and revisions are easily done with the monitor. If you have the Super Expansion Board and Super Monitor the monitor is up and running at the push of a button.

Other board options include Parallel Input and Output Ports with full handshake. They allow easy connection of an ASCII keyboard to the input port. RS 232 and 20 ma Current Loop for teletype or other device are on board and if you need more memory there are two S-100 slots for static RAM or video boards. A Godbout 8K RAM board is available for \$135.00. Also a 1K Super Monitor version 2 with video driver for full capability display with Tiny Basic and a video interface board. Parallel I/O Ports \$9.95, RS 232 \$4.50, 20 ma I/F \$1.95, S-100 \$4.50. A 50 pin connector set with ribbon cable is available at \$12.50 for easy connection between the Super Elf and the Super Expansion Board.

The Power Supply Kit for the Super Expansion Board is a 5 amp supply with multiple positive and negative voltages \$29.95. Add \$4.00 for shipping. Pre-punched frame \$7.50. Case \$10.00. Add \$1.50 for shipping.

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8v 5 amp, ±18v 5 amp, 5v 1.5 amp, -5v 5 amp, 12v 5 amp, -12v option, ±5v, ±12v are regulated. Kit \$29.95. Kit with punched frame \$37.45. Woodgrain case \$10.00

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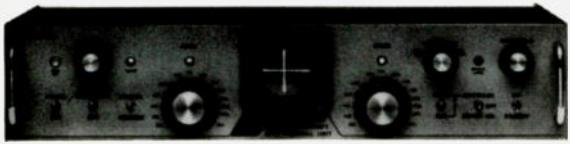
INTEGRATED CIRCUITS

7400TTL	LW3177/K	2.92	CD4015	86				
7400N	17	LW318	1.35	CD4016	36			
7420N	17	LW320K-5	1.20	CD4017	94			
7404N	19	LW320K-4	6.95	CD4018	28			
7402N	23	LW320K-3	1.20	CD4019	28			
7410N	17	LW320K-15	1.35	CD4020	1.02	2102-1	3.95	
7414N	63	LW320T-5	1.60	CD4021	1.02	2102-1	3.95	H4025136
7420N	19	LW320T-4	1.60	CD4022	86	2102-4	1.18	H4025137
7404N	19	LW320T-3	1.60	CD4023	78	2102-3	1.18	2708
7430N	20	LW320T-15	1.60	CD4024	75	2102	1.25	D40257
7442N	45	LW324	1.55	CD4025	28	2104-4	4.95	S223
7454N	69	LW339N	1.55	CD4026	1.51	2104-4	2.90	2716 inal
7448N	69	LW340K-8	1.35	CD4027	2113-1		3.75	2716 inal
7450N	17	LW340K-12	1.35	CD4029	1.02	2114L-3	3.95	
7451N	29	LW340K-12	1.35	CD4030	78	2114L-3	3.95	
7475N	49	LW340T-12	1.35	CD4035	1.02	2114-2	6.30	44 pin edge
7485N	88	LW340T-5	1.25	CD4040	1.02	M40526	2.90	100 pin edge
7489N	200	LW340T-8	1.25	CD4042	71	M40526	3.00	100 pin edge WV 5.25
7491N	43	LW340T-15	1.25	CD4044	63	M40530	9.95	
7493N	43	LW340T-15	1.25	CD4045	1.67	PD411D-3	5.94	
7495N	69	LW340T-24	1.25	CD4049	36	PD411D-3	5.94	
7497N	90	LW340T-24	1.25	CD4050	36	PD411D-3	5.94	
7410T	17	LW350	7.50	CD4051	1.13	4200A	9.95	14
7412T	34	LW350	1.50	CD4060	7.90	4200A	9.95	18
7413T	59	LW357	3.00	CD4061	71	4200B	9.95	20
7414T	69	LW357	1.50	CD4062	1.02	H00165-5	6.95	20
7415T	95	LW381	1.60	CD4070	40	M405100	9.95	24
7415T	151	LW381	1.60	CD4071	39	M405100	9.95	24
7415T	151	LW381	1.60	CD4072	39	M405100	9.95	24
7415T	69	LW381	2.80	CD4073	40	M405100	9.95	24
7416T	87	LW723H/N	5.00	CD4075	28	416	16.00	WIRE WRAP LEVEL 3
7416T	87	LW723N	5.00	CD4075	1.75		14	25
7417T	87	LW723N	5.00	CD4076	1.75		16	33
7417T	87	LW723N	5.00	CD4077	1.75		16	38
7417T	87	LW723N	5.00	CD4078	1.75		16	40
7417T	87	LW723N	5.00	CD4079	1.75		16	42
7417T	87	LW723N	5.00	CD4080	1.75		16	44
7417T	87	LW723N	5.00	CD4081	1.75		16	46
7417T	87	LW723N	5.00	CD4082	1.75		16	48
7417T	87	LW723N	5.00	CD4083	1.75		16	50
7417T	87	LW723N	5.00	CD4084	1.75		16	52
7417T	87	LW723N	5.00	CD4085	1.75		16	54
7417T	87	LW723N	5.00	CD4086	1.75		16	56
7417T	87	LW723N	5.00	CD4087	1.75		16	58
7417T	87	LW723N	5.00	CD4088	1.75		16	60
7417T	87	LW723N	5.00	CD4089	1.75		16	62
7417T	87	LW723N	5.00	CD4090	1.75		16	64
7417T	87	LW723N	5.00	CD4091	1.75		16	66
7417T	87	LW723N	5.00	CD4092	1.75		16	68
7417T	87	LW723N	5.00	CD4093	1.75		16	70
7417T	87	LW723N	5.00	CD4094	1.75		16	72
7417T	87	LW723N	5.00	CD4095	1.75		16	74
7417T	87	LW723N	5.00	CD4096	1.75		16	76
7417T	87	LW723N	5.00	CD4097	1.75		16	78
7417T	87	LW723N	5.00	CD4098	1.75		16	80
7417T	87	LW723N	5.00	CD4099	1.75		16	82
7417T	87	LW723N	5.00	CD4100	1.75		16	84
7417T	87	LW723N	5.00	CD4101	1.75		16	86
7417T	87	LW723N	5.00	CD4102	1.75		16	88
7417T	87	LW723N	5.00	CD4103	1.75		16	90
7417T	87	LW723N	5.00	CD4104	1.75		16	92
7417T	87	LW723N	5.00	CD4105	1.75		16	94
7417T	87	LW723N	5.00	CD4106	1.75		16	96
7417T	87	LW723N	5.00	CD4107	1.75		16	98
7417T	87	LW723N	5.00	CD4108	1.75		16	100
7417T	87	LW723N	5.00	CD4109	1.75		16	102
7417T	87	LW723N	5.00	CD4110	1.75		16	104
7417T	87	LW723N	5.00	CD4111	1.75		16	106
7417T	87	LW723N	5.00	CD4112	1.75		16	108
7417T	87	LW723N	5.00	CD4113	1.75		16	110
7417T	87	LW723N	5.00	CD4114	1.75		16	112
7417T	87	LW723N	5.00	CD4115	1.75		16	114
7417T	87	LW723N	5.00	CD4116	1.75		16	116
7417T	87	LW723N	5.00	CD4117	1.75		16	118
7417T	87	LW723N	5.00	CD4118	1.75		16	120
7417T	87	LW723N	5.00	CD4119	1.75		16	122
7417T	87	LW723N	5.00	CD4120	1.75		16	124
7417T	87	LW723N	5.00	CD4121	1.75		16	126
7417T	87	LW723N	5.00	CD4122	1.75		16	128
7417T	87	LW723N	5.00	CD4123	1.75		16	130
7417T	87	LW723N	5.00	CD4124	1.75		16	132
7417T	87	LW723N	5.00	CD4125	1.75		16	134
7417T	87	LW723N	5.00	CD4126	1.75		16	136
7417T	87	LW723N	5.00	CD4127	1.75		16	138
7417T	87	LW723N	5.00	CD4128	1.75		16	140
7417T	87	LW723N	5.00	CD4129	1.75		16	142
7417T	87	LW723N	5.00	CD4130	1.75		16	144
7417T	87	LW723N	5.00	CD4131	1.75		16	146
7417T	87	LW723N	5.00	CD4132	1.75		16	148
7417T	87	LW723N	5.00	CD4133	1.75		16	150
7417T	87	LW723N	5.00	CD4134	1.75		16	152
7417T	87	LW723N	5.00	CD4135	1.75		16	154
7417T	87	LW723N	5.00	CD4136	1.75		16	156
7417T	87	LW723N	5.00	CD4137	1.75		16	158
7417T	87	LW723N	5.00	CD4138	1.75		16	160
7417T	87	LW723N	5.00	CD4139	1.75		16	162
7417T	87	LW723N	5.00	CD4140	1.75		16	164
7417T	87	LW723N	5.00	CD4141	1.75		16	166
7417T	87	LW723N	5.00	CD4142	1.75		16	168
7417T	87	LW723N	5.00	CD4143	1.75		16	170
7417T	87	LW723N	5.00	CD4144	1.75		16	172
7417T	87	LW723N	5.00	CD4145	1.75		16	174
7417T	87	LW723N	5.00	CD4146	1.75		16	176
7417T	87	LW723N	5.00	CD4147	1.75		16	178
7417T	87	LW723N	5.00	CD4148	1.75		16	180
7417T	87	LW723N	5.00	CD4149	1.75		16	182
7417T	87	LW723N	5.00	CD4150	1.75		16	184
7417T	87	LW723N	5.00	CD4151	1.75		16	186
7417T	87	LW723N	5.00	CD4152	1.75		16	188
7417T	87	LW723N	5.00	CD4153	1.75		16	190
7417T	87	LW723N	5.00	CD4154	1.75		16	192
7417T	87	LW723N	5.00	CD4155	1.75		16	1

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Standard features include CONTINUOUSLY tuneable Mark and Space channels (1000 Hz to 3200 Hz), Dual Mode (MARK or FSK) Autostart and internal high level neutral loop keyer (20 to 60 ml). Both EIA and MIL FSK outputs are provided for direct interface to microprocessor and video terminal peripherals.



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A front panel switch permits internal TSR-200 Signal Regenerator-Speed converter assembly to electronically "gear-shift" between 60, 67, 75 and 100 WPM. All incoming and outgoing signals are regenerated to less than 0.5% bias distortion. Also available with DIGITAL Autostart (TSR-200D): Amateur Net: \$695.00



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Dual UART Regeneration,
Speed Conversion, 200
Char. Memory, Word Cor-
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Autostart

Amateur Net: \$895.00*

The MPC-1000R/TSR-500 provides Preloading and Recirculation of the 200 character FIFO Memory, a keyboard-controlled Word Correction circuit, Variable Character Rate, Tee Dee Inhibit, Blank/LTRS Diddle, a Triple Tone-Pair AFSK Tone Keyer and a Character Recognition/Speed Determination DIGITAL (DAS-100) Autostart mode.

*The MPC-1000R is also available without a TSR assembly and functions as a MPC-1000C with a Triple Tone-Pair AFSK Tone Keyer. This "Basic-R" permits future expansion with a TSR-100, TSR-200, TSR-200D or TSR-500 by simply lifting the lid and plugging in the appropriate TSR assembly: Amateur Net (Basic-R): \$595.00

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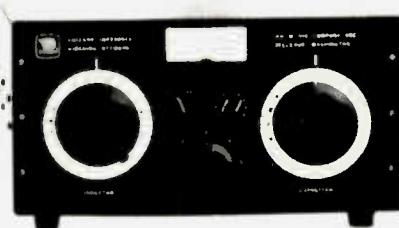
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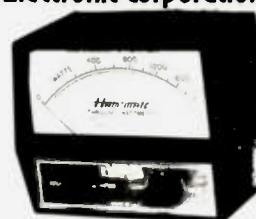
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100 watts	100.0
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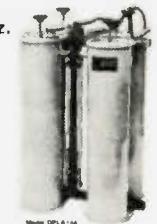
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- RIT
- RF ATT
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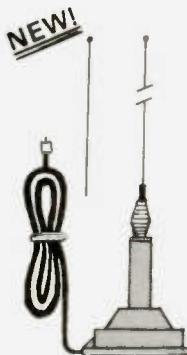
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4-1000A	255.00	6146A	5.25
5-500A	145.00	6146B	6.25
4CX250B	38.50	6159	10.60
4CX250F	53.50	6293	18.50
4-125A	68.75	811A	12.95
4-250A	80.00	813	29.00
4-400A	81.50	5894	39.00
4-1000A	255.00	6146A	5.25
5-500A	145.00	6146B	6.25
4CX250B	38.50	6159	10.60
4CX250F	53.50	6293	18.50
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4-400A	81.50	5894	39.00
4-1000A	255.00	6146A	5.25
5-500A	145.00	6146B	6.25
4CX250B	38.50</td		

MURATA CERAMIC FILTERS				19.2 KC	ALL CRYSTALS \$4.95					
					2.2395 MC	2.854 MC	3.4045 MC	6.753 MC	20.1MC	21.99965
Model SFD-455D		31.5								
455 kHz \$3.00		37.35		2.24075	2.854285	3.4115	6.7562	22.		
Model SFB-455D		49.710		2.241	2.865	3.4325	6.7605	23.25		
455 kHz \$2.00		70		2.246	2.868	3.4535	6.7712	23.575		
Type: CFM-455E 455 kHz	\$7.95	81.9		2.2475	2.8725	3.4675	6.77625	25.47667		
Type: SFE-10.7 10.7 MHz	\$5.95	96		2.2925	2.876875	3.4815	6.880000	25.9		
225		2.2975		2.887	3.5	6.910	25.99961			
TEKTRONIX		250		2.320	2.889	3.579545	6.940	26.66667		
585A Oscilloscope		285.714		2.326	2.894	3.64	7.0057	26.8965		
with 82 dual trace plug-in		576		2.32625	2.910	3.7735	7.15	26.9		
DC-100 MHz	\$800.00	720		2.32885	2.920	3.80	7.35	26.958		
HEWLETT PACKARD		1.0000MC		2.3525	2.925450	3.805	7.390	27.77778		
Model 140A Oscilloscope		1.2288		2.35256	2.92545	3.803	7.443	27.9		
W/1402A, 1400A	\$495.00	(also available 1415A \$300)		1.3047	2.368	2.931	3.901	7.473		
TEKTRONIX		1.4		2.374	2.94375	3.908	7.5	28.728		
IL30 Spectrum Analyzer plug-in		1.455		2.375	2.945	3.9168	7.81	28.9		
925MHz-10.25GHz	\$899.00	1.689600		2.38725	2.94875	4.0000	8.00764	28.93888		
1.7		2.395		2.952	4.011	8.00824	29.896			
1.76375		2.396875		2.966	4.126666	8.075	29.9			
HEWLETT PACKARD UHF, VHF, AND MICROWAVE SIGNAL		1.77125		2.42	2.973	4.26	8.12	30.0000		
GENERATORS AND SWEEPERS, AND OTHER EQUIPMENT		1.773125		2.4375	2.980	4.3	8.15571	30.9		
MODEL 434A		1.78675		2.44275	2.981	4.57	8.364	31.0000		
Calorimetric power meter	\$450.00	1.80224		2.4495	2.98325	4.6895	8.820	31.11111		
Ratio meter	\$125.00	1.81875		2.45	2.987	4.6965	8.8285	31.66667		
DC null voltmeter	\$112.50	1.845125		2.4585	2.9989	4.7	8.837	31.9		
1.84375		2.46125		3.001	4.7175	8.8455	32.0000			
MODEL 400DR		1.845625		2.482	3.0235	4.7245	8.854	32.22222		
Vacuum tube voltmeter	\$79.95	1.84575		2.486	3.045	4.7315	8.8625	32.9		
1.846		2.5		3.049	4.765	8.871	33.0000			
A Model \$299.00		1.8425		2.51375	3.053	4.89	8.879500	33.33333		
1.84975		2.56		3.062	5.0000	8.888	33.9			
MODEL 606A		1.8575		2.581	3.067	5.13125	8.905	34.0000		
50 kHz to 65 MHz		1.908125		2.604	3.074	5.139585	8.9305	34.44444		
.1mV to 3V into 50 ohms	ONLY \$299.00	1.925		2.6245	3.1125	5.147917	8.939	34.44444		
\$1,000.00		1.927		2.618	3.126	5.164583	8.956	35.0000		
1.932		2.62825		3.137	5.348400	9.0265	35.25000			
MODEL TS510/U HP608D		1.982		2.633125	3.13975	5.426636	9.65	35.55555		
10 MHz to 420 MHz		1.985		2.639	3.1435	5.436636	9.65	36.0000		
.1V to .5V	.223V to 1uv.	1.9942		2.63575	3.144	5.456	9.7	36.21750		
\$399.95	\$699.99	1.995975		2.64325	3.145	5.4675	9.75	36.66667		
302A with a 297A		1.964750		2.646	3.151	5.4990	9.8	37.00000		
7 to 11 GHz		2.0000		2.647	3.1545	5.5065	9.85	37.2175		
Wave Analyzer and Sweep Drive		2.0285		2.650750	3.158	5.515	9.9	37.385		
20 HZ to 50 kHz		2.03505		2.6545	3.1585	5.5215	9.95	37.460		
ONLY \$9.95 or 2/\$18.00		2.126175		2.65825	3.1615	5.544	9.99	37.77777		
2.12795		2.660		3.1625	5.5515	10.0000	38.00000			
2.1315		2.662		3.166	5.559	10.021	38.33333			
TRW BROADBAND AMPLIFIER MODEL CA615B		2.133275		2.66575	3.16975	5.5665	10.20833	38.77777		
Frequency response 40 to 300 MHZ		2.13505		2.6695	3.177	5.574	10.80375	38.77778		
Gain		2.136825		2.677	3.181	5.5815	11.	38.88888		
300 MHZ 16dB MIN.		2.1425		2.68075	3.1825	5.589	11.1805	39.00000		
17.5dB MAX.		2.144625		2.681	3.18475	5.604	11.228	39.160		
50 MHZ 0 to -1dB from 300 MHZ		2.14675		2.6845	3.1885	5.619	11.2375	40.00000		
Voltage		2.148875		2.68825	3.2035	5.6115	11.2995	41.11111		
24 volts DC at 220ma MAX		2.151		2.69575	3.20725	5.6265	11.3565	41.66666		
ONLY \$14.95		2.153125		2.702	3.2105	5.6415	11.535	43.33333		
CIRCUIT BOARD DRILL BITS		2.15375		2.704	3.2175	5.6715	11.69626	45.		
Size	Price	Size	Price							
35	\$2.15	58	\$1.85	2.155		5.675	12.49	47.48		
42	\$2.15	59	\$1.85	2.15525		5.680	12.99	49.84166		
47	\$2.15	60	\$1.85	2.157375		5.695	13.09	49.95		
49	\$2.15	61	\$1.85	2.1595		5.7	13.102	53.45		
50	\$2.15	62	\$1.85	2.16375		5.723	13.2155	54.95		
51	\$2.15	63	\$1.85	2.165875		5.73775	13.2455	57.45		
52	\$2.15	64	\$1.85	2.170125		5.7315	13.2845	59.45		
53	\$1.85	65	\$1.85	2.17225		5.73225	13.28875	61.45		
54	\$1.85	66	\$1.90	2.174375		5.732625	13.2925	62.10		
55	\$1.85	1.25 mm	\$1.85	2.1765		5.733	13.3245	63.80833		
56	\$1.85	1.45 mm	\$1.85	2.17925		5.737	13.3445	66.66667		
57	\$1.85	3.20 mm	\$3.58	2.18475		5.74025	13.345	72.855		
				2.18575		5.742125	13.345	75.185		
				2.194125		5.7425	13.345	76.66667		
INTEGRATED CIRCUITS		2.207063		5.744		5.744	13.3545	82.75		
MC1303L	\$ 2.00	MC1460R	\$ 5.40	2.208313		5.7445	14.315	83.		
MC1461R	6.90	MC1463R	5.15			5.24975	15.016	84.		
MC1469G	2.05	MC1469R	3.55	2.209563		5.4975	15.036	85.833330		
MC1550G	1.50	MC1560G	10.20	2.210812		5.4975	15.067	90.833		
MC1560R	12.40	MC1563R	10.00	2.210813		5.5215	15.2800	93.1346		
MC1568G	5.31	MC1568L	5.00	2.210263		5.525	15.612	97.8710		
MC1569R	8.15	MC1590G	6.50	2.214562		5.5645	17.9065	99.9353		
MC4204P	3.82	MC8800P	9.95	2.214563		5.58625	17.9165	94.3		
MC6820P	6.95	MC68B21P	12.00	2.215625		5.673	17.9265	95.		
2513	6.95	4116-200NS	10.37	2.217938		5.723	17.9365	95.35		
8080A	3.95	TMS4080	6.95	2.21975		5.726625	17.9465	100.0		
2708TI	8.95	TMS4024	13.90	2.222125		5.731125	17.9665	106.850		
2716TI	29.95	1702A	4.95	2.22325		5.732625	17.9735	123.5		
				2.22675		5.736125	17.9735	146.64		
				2.22875		5.7375	17.9735	147.09		
				2.23725		5.7375	17.9735	165.5		

Prices are subject to change. Some items are in limited quantity.



FREQUENCY COUNTER KIT

Outstanding Performance

CT-50

Incredible Price

\$89.95

The CT-50 is a versatile and precision frequency counter which will measure frequencies to 60 mHz and up to 600 mHz with the CT-600 option. Large Scale Integration, CMOS circuitry and solid state display technology have enabled this counter to match performance found in units selling for over three times as much. Low power consumption (typically 300-400 ma) makes the CT-50 ideal for portable battery operation. Features of the CT-50 include: large 8 digit LED display, RF shielded all metal case, easy pushbutton operation, automatic decimal point, fully socketed IC chips and input protection to 50 volts to insure against accidental burnout or overload. And, the best feature of all is the easy assembly. Clear, step by step instructions guide you to a finished unit you can rely on.

Order your today!

CT-50, 60 mHz counter kit

CT-50WT, 60 mHz counter, wired and tested

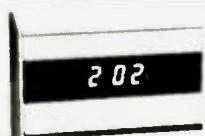
CT-600, 600 mHz scaler option, add

\$89.95
159.95
29.95

CB-1, Color TV calibrator-stabilizer
DP-1, DC probe, general purpose probe
HP-1, High impedance probe, non-loadin

\$14.95
12.95
15.95

CAR CLOCK



The UN-KIT, only 5 solder connections

Here's a super looking rugged and accurate auto clock which is a snap to build and install. Clock movement is completely assembled—you only solder 3 wires and 2 switches. Takes about 15 minutes! Display is bright green with automatic brightness control photocell—assures you of a highly readable display, day or night. Comes in a satin finish anodized aluminum case which can be attached 5 different ways using 2 sided tape. Choice of silver, black or gold case (specify).

DC-3 kit, 12 hour format \$22.95
DC-3 wired and tested \$29.95
110V AC adapter \$5.95

Under dash car clock



12/24 hour clock in a beautiful plastic case features 6 jumbo RED LEDs, high accuracy (1min/mo.), easy 3 wire hookup, display blinks with ignition, and super instructions. Optional dimmer automatically adjusts display to ambient light level.

DC-11 clock with mtg. bracket \$27.95
DM-1 dimmer adapter 2.50

PRESCALER

Extend the range of your counter to 600 mHz. Works with any counter. Includes 2 transistor pre-amp to give super sens. typically 20 mv at 150 mHz. Specify -10 or +100 ratio. PS-1B, 600 mHz prescaler \$59.95
PS-1BK, 600 mHz prescaler kit 49.95



Ramsey's famous MINI-KITS

FM WIRELESS MIKE KIT

Transmits up to 300' to any FM broadcast radio; uses any type of mike. Runs on 3 to 9V. Type FM-2 has added sensitive mike preamp stage. FM-1 kit \$2.95 FM-2 kit \$4.95



VIDEO MODULATOR KIT

Converts any TV to video monitor. Super stable, tunable over ch 4-6. Runs on 5-15V, accepts std. video signal. Best unit on the market! Complete kit, VD-1 \$6.95

TONE DECODER

A complete tone decoder on a single PC board. Features 400-5000 Hz adjustable range via 2 turn pot, voltage regulation, 567 IC. Useful for touch-tone decoding, tone burst detection, FSK etc. Can also be used as a stable tone encoder. Runs on 5 to 12 volts. Complete kit, TD-1 \$5.95



COLOR ORGAN/MUSIC LIGHTS

See music come alive! 3 different lights flicker with music. One light for lows, one for the mid-range and one for the highs. Each channel individually adjustable and drives up to 300W. Great for parties, band music, night clubs and more! Complete kit, ML-1 \$7.95

LED BLINKY KIT

A great attention getter which alternately flashes 2 jumbo LEDs. Use for name badges, buttons, warning panel lights, anything! Runs on 3 to 15 volts. Complete kit, BL-1 \$2.95

WHISPER LIGHT KIT

An interesting kit, small mike picks up sounds and converts them to light. The louder the sound, the brighter the light. Completely self-contained, includes mike, runs on 110VAC, controls up to 300 watts. Complete kit, WL-1 \$6.95

SUPER SLEUTH

A super sensitive amplifier which will pick up a pin drop at 15 feet! Great for monitoring baby's room or as general purpose amplifier. Full 2 W rms output, runs on 6 to 15 volts, uses 8-45 ohm speaker. Complete kit, BN-9 \$5.95



POWER SUPPLY KIT

Complete triple regulated power supply provides variable .6 to 18 volts at 200 mA and +5V at 1 A. Amp. Excellent load regulation, good filtering and small size. Less transformers, requires 6.3V, 1 A and 24 VCT. Complete kit, PS-3LT \$6.95



SIREN KIT

Produces upward and downward wall characteristic of a police siren. 5W peak audio output runs on 3-15 volts, uses 3-45 ohm speaker. Complete kit, SM-3 \$2.95



ramsey electronics

BOX 4072, ROCHESTER, N.Y. 14610

PHONE ORDERS CALL
(716) 271-6487



TERMS: Satisfaction guaranteed or money refunded. COD, add \$1.50. Minimum order, \$6.00. Orders under \$10.00, add \$7.75. Add \$5 for postage, insurance, handling. Overseas, add 15% NY residents, add 7% tax.

These Low Cost SSB TRANSMITTING CONVERTERS

Let you use inexpensive recycled 10 or 11 meter SSB excitors on VHF!



- Linear Converter for SSB, CW, FM, etc.
- A fraction of the price of other units
- 2W p.e.p. output with 1 mW drive
- Use low power tap on exciter or attenuator pad
- Easy to align with built-in test points
- Link with VHF RX converter for transceive

KIT ONLY \$69.95

MODEL	INPUT (MHz)	OUTPUT (MHz)
XV2-1	28-30	50-52
XV2-2	28-30	220-222
XV2-3	28-30 (26-28)	222-224 (220-222)
XV2-4	28-30	144-146
XV2-5	28-30	145-147
XV2-6	26-28	144-146
XV2-7	144-146	50-52
XV2-8	144-146	220-222

FAMOUS HAMTRONICS PREAMPS

Let you hear the weak ones too!
Great for OSCAR, SSB, FM, ATV. Over 14,000 in use throughout the world on all types of receivers.



P9 Kit \$12.95
P14 Wired \$24.95
Specify band when ordering

- Deluxe vhf model for applications where space permits
- 1 1/2" x 3"
- Models available to cover any 4 MHz band in the 26 to 230 MHz range
- 12 Vdc
- 2 stages
- Ideal for OSCAR
- 20 dB gain

P8 Kit \$10.95
Specify band when ordering

- Miniature vhf model for tight spaces—size only 1/2 x 2 1/2"
- Models available to cover any 4 MHz band in the range 20 to 230 MHz
- 20 db gain
- 12 Vdc

P15 Kit \$18.95
P35 Wired \$34.95



- Covers any 6 MHz band in UHF range of 380 to 520 MHz
- 20 dB gain
- 2 stages
- 12 Vdc

IT'S EASY TO ORDER!

- Write or phone 716-392-9430
(Electronic answering service evenings & weekends)
- Use Credit Card, UPS COD, Check, Money Order
- Add \$2.00 shipping & handling per order

Easy to Build FET RECEIVING CONVERTERS

Let you receive OSCAR and other exciting VHF and UHF signals on your present HF or 2M receiver

VHF KIT STILL ONLY \$34.95



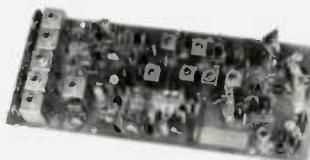
MODEL	RF RANGE	OUTPUT RANGE
C28	28-32 MHz	144-148
C50	50-52	28-30
C50-2	50-52	144-146
C144	144-146	28-30
C145	145-147	28-30
C146	146-148	28-30
C146	144-146	26-28
C220	220-222	28-30
C220-2	220-222	144-146
C110	Any 2 MHz of Aircraft Band or 28-30	26-28
C110-ELT	121.5 (121.6)	CB Chan 9 (17)

UHF KIT ONLY \$34.95

MODEL	RF RANGE	OUTPUT RANGE
C432-2	432-434	28-30
C432-5	435-437	28-30
C432-4	432-436	144-148
C432-7	427.25	61.25
C432-9	439.25	61.25

Professional Quality VHF/UHF FM/CW EXCITERS

- Fully shielded designs
- Double tuned circuits for spurious suppression
- Easy to align with built-in test aids



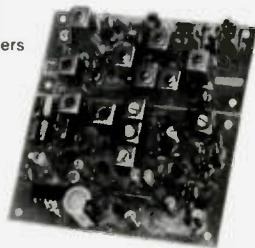
T50-50	6-chan, 6M, 2W Kit	\$49.95
T50-150	6-chan, 2M, 2W Kit	\$49.95
T50-220	6-chan, 220 MHz, 1 1/2 W Kit	\$49.95
T40/T20	11-chan, 450 MHz, 200mW Kit	\$49.95

See our Complete Line of VHF & UHF Linear PA's

- Use as linear or class C PA
 - For use with SSB Xmtg Converters, FM Exciters, etc.
- | | | |
|----------|--|----------|
| LPA 2-15 | VHF PA, 2W in/15-20W out. Solid-state t/r switching. Kit only | \$59.95 |
| LPA 2-45 | VHF PA, 2W in/40-45W out. Can also be used with 8-10W drive. Kit price | \$109.95 |
| LPA 4-10 | UHF PA, 200-500mW in/6-10W out. Kit price only | \$79.95 |

New R75 One Channel VHF FM RECEIVER Offers Unprecedented Range of Selectivity Options!

- New generation
- More sensitive
- More selective
- Uses crystal filters
- Smaller
- Easy to align



R75A Kit for monitor or weather satellite service. -60dB at ± 30 KHz.....	\$69.95
R75B Kit for normal nbfm service. -60dB at ± 17 KHz, -80dB at ± 25 KHz.....	\$74.95
R75C Kit for repeater service. -60dB at ± 14 KHz, -80dB at ± 22 KHz.....	\$84.95
R75D Kit for split channel operation. Uses 8 pole crystal filter! -60dB at ± 9 KHz, -100dB at ± 15 KHz...\$99.95	

Specify band: 10M, 6M, 2M, or 220 MHz. May also be used on adjacent commercial bands. Use 2M version for 137MHz WX satellite.

HAMTRONICS SIX CHANNEL VHF & UHF FM RECEIVERS

In use by the hundreds throughout the world. Unlimited applications.

- Commercial grade design
- Easy to build & align
- 70 or 100dB selectivity options
- Low system cost
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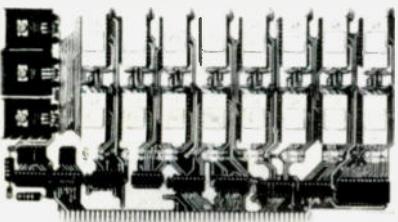
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USES 2708's!

Thousands of personal and business systems around the world use this board with complete satisfaction. Puts 16K of software on line at **ALL TIMES!** Kit features a top quality soldermasked and silk-screened PC board and first run parts and sockets. All parts (except 2708's) are included. Any number of EPROM locations may be disabled to avoid any memory conflicts. Fully buffered and has WAIT STATE capabilities.

OUR 450NS 2708'S
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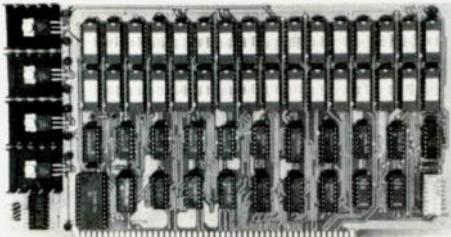
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ADD \$25

16K STATIC RAM KIT-S 100 BUSS

\$279 KIT

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FULLY
STATIC, AT
DYNAMIC PRICES



WHY THE 2114 RAM CHIP?

We feel the 2114 will be the next industry standard RAM chip (like the 2102 was). This means price, availability, and quality will all be good! Next, the 2114 is **FULLY STATIC!** We feel this is the **ONLY** way to go on the S-100 Bus! We've all heard the HORROR stories about some Dynamic Ram Boards having trouble with DMA and FLOPPY DISC DRIVES. Who needs these kinds of problems? And finally, even among other 4K Static RAM's the 2114 stands out! Not all 4K static Rams are created equal! Some of the other 4K's have clocked chip enable lines and various timing windows just as difficult as Dynamic RAM's. Some of our competitor's 16K boards use these "tricky" devices. But not us! The 2114 is the **ONLY** logical choice for a trouble-free, straightforward design

KIT FEATURES

- 1 Addressable as four separate 4K Blocks
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- 3 Uses 2114 (450NS) 4K Static Rams
- 4 ON BOARD SELECTABLE WAIT STATES
- 5 Double sided PC Board, with solder mask and silk screened layout. Gold plated contact fingers
- 6 All address and data lines fully buffered
- 7 Kit includes ALL parts and sockets
- 8 PHANTOM is jumpered to PIN 67
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BLANK PC BOARD W/DATA-\$33

LOW PROFILE SOCKET SET-\$12
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2114 RAM'S—8 FOR \$69.95

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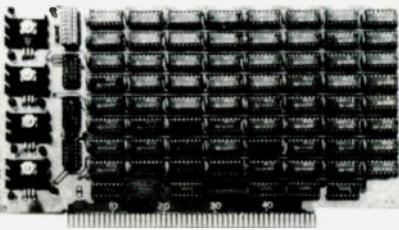
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- 2 All sockets included.
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- 5 FOUR 7805 regulators are provided on card.

Blank PC Board w/Documentation

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Low Profile Socket Set...13.50

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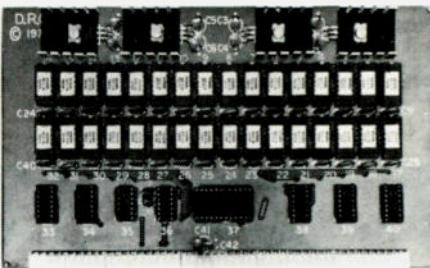
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- 5 All Parts and Sockets included
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NEW!

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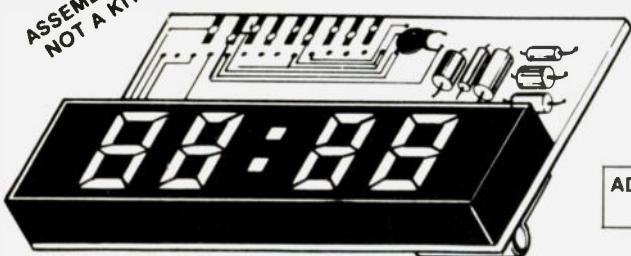
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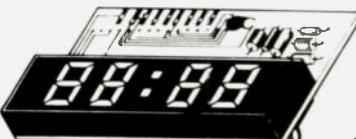
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5% x 10% In. 120 VAC, 120 WATTS. Made of 1/4 In. tempered plate glass with Ni-Chrome heating element laminated to back. Element size is 4 1/4 x 9 1/4 Inches. Double protected by TIKLIXON Thermostat and two thermal fuses. Each also has neon ready light. Besides the obvious use as a bun warmer, food warmer, coffee warmer, glue warmer, etc., our tests show this plate to be an excellent warmer for ferric chloride solution used in etching PC Boards by hobbyists. Typically increases etching efficiency by 300% over room temperature. Non-Submersible.

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- One half the length of conventional half wave dipoles
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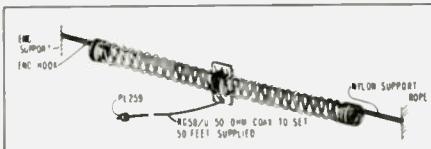
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 - 50' 8U foam coax.
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\$59.95

Wt. 16 lbs.

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• Wide band performance	
• New Hy-Q traps	
Self-supporting, automatic band switching vertical antenna Omni-directional performance Favorable L/C ratio. High Q. True 1/4 wave resonance on all bands. Low angle radiation pattern. Taper swaged seamless aluminum construction 12" double grip mast bracket Full circumference compression clamps at tubing joints Weight 8.2 lbs	
Model 14AVQ/WB	\$57.00

Model 18AVT/WB	\$79.95
• Automatic band switching	
• Completely Self-Supporting	
• Omni-Directional Performance	
Three beeld-up Hy-Q traps permit automatic switching 5 band capability. Favorable L/C ratio Top loading coil Across the band performance with one furnished setting for each band (10 thru 40). True 1/4 wave resonance on all bands. SWR of 2:1 or less at band edges. Low angle radiation pattern. Extra heavy duty tapered swaged seamless aluminum tubing with full circumference. corrosion resistant compression clamps at tubing joints. Antenna can be mounted without guide wires. 25 high. Weight 10.7 lbs.	
Model 18 AVT/WB	\$79.95

BN86 Ferrite balun for 80-10mtrs....	\$15.95 a
155 Center insulator for doublet.....	5.95 a
156 End insul. for doublet (pair)	3.95 a
18HT ★ HyTower 80-10M vertical	239.95 ★
18V Economy 80 tru 10M vertical	24.95 c
12AVQ 20-10mtr trap vertical	39.95 c
14AVQ 40-10mtr trap vertical	57.00 d
18AVT/WB 80-10mtr trap vertical	79.95 d
2BDQ Trap doublet for 80 & 40mtr	39.95 d
5BDQ Trap doublet for 80-10mtrs....	69.95 e
TH3 MkIII ★ 3 ele 20-10 tribander	179.95 ★
TH6DX ★ 6 ele 20-10M tribander	239.95 ★
TH3 Jr 3 ele tribander (750W PEP)	129.95 e
HY QUAD ★ 2 ele quad 20-10 mtrs	189.95 ★
103BA 3 element 10Mtr beam	54.95 e
153BA ★ 3 element 15mtr beam	79.95
204BA ★ 4 element 20mtr beam	179.95 ★
402BA ★ 2 element 40 mtr beam	169.95 ★
64B ★ 4 element 6 meter beam	39.95
270 6db fiberglass 2M antenna	39.95 e
203 3 element 2 meter yagi	15.95 b
205 5 element 2 meter yagi	17.95 c
208 8 element 2 meter yagi	25.95 c
214 14 element 2 meter yagi	31.95 d

Mosley

TA33 ★ 3 el triband beam 2KW PEP ..	\$189.95
TA33Jr ★ 3ele triband 1KW PEP	149.95 ★
TA36 ★ 6ele triband 2KW PEP	269.95 ★
CL33 ★ 3 ele Classic	209.95 ★
CL36 ★ 6 ele Classic	269.95 ★
TA40KR 40mtr adapt for TA33/36	92.95 e
MPK3 2KW conv kit for TA33Jr	67.65 e
CDE	
HAM IV Rotor with control box	\$149.95 f
T2X Tailtwister xtra hvy dty	219.95 f
CD45 Medium duty rotor	119.95 f
AR22XL Light duty rotor	59.95 e
Mast adaptor for Ham II/T2X	29.95 d
South Center meter kit	3.00 a

THE SUMMER SPECIALS CONTINUE !!!

★ 16K MEMORY EXPANSION CHIP SET ★ \$87.20 ★

Expand memory in the Apple, Radio Shack-80, and Exidy Sorcerer computers — for 20% off our regular price! Why is this kit one of our all time best sellers? Many reasons... 250 ns access time chips for use with 4 MHz systems, low power ICs used exclusively, DIP shunts included, 1 year limited warranty, and easy-to-follow instructions that make memory expansion a snap. We've not only made it simple for you to acquire a more powerful memory, we've made it economical too.

MA1003 CLOCK MODULE ★ \$13.20 ★

Our very best clock module is reduced by 20% for another month... but don't count on these lasting too much longer. Features include an internal crystal timebase accurate to .01%, fluorescent readouts that don't wash out in daylight (and look beautiful at night!), simple assembly (just add 2 time-setting switches and 12V DC), and much more. Our applications note tells you how to use this module to maximum advantage in your car, van, boat, or airplane. Also available: matching case with mounting hardware and optical filter for \$5.95.

Special prices good through cover month of magazine.
All items are limited quantity — first come, first served.

HAMS! ON RF POWER ★ 10% OFF TRANSISTORS ★

2NRF-1	2 GHz RF power transistor. Pd max (@ 25 degrees C) 3.5W, Pout min @ 2 GHz 1.0W, Pin 310 mW, efficiency @ 2 GHz 30%, round shape, similar to RCA 2N5470.	\$4.45
2NRF-2	2 GHz RF power transistor. Pd 8.7W, Pout 2.5W, Pin 300 mW, efficiency 33%, cross shape, similar to RCA TA8407.	\$5.35
2NRF-3	2 GHz RF power transistor. Pd 21W, Pout 5.5W, Pin 1.25W, efficiency 33%, cross shape. Similar to RCA 2N6269.	\$6.25
2NRF-4	2 GHz RF power transistor. Pd 29W, Pout 7.5W, Pin 1.5W, efficiency 33%, cross shape. Factory selected, prime 2N6269.	\$7.15

★ WE BREAK THE 1C/BYTE BARRIER !

We're offering low power 21L02 1K static RAM memory chips, guaranteed to run with any 2 MHz system, at the very low price of 10/\$9.90... that's under 1¢ per byte. Stock up now, we can't predict how much longer we'll have the last of these prime parts available for sale.

GODBOU SLASHES STATIC MEMORY PRICES AGAIN: S-100 32K \$529, 24K \$398, 16K \$269!

Econoram* unkits are now at their lowest prices ever. What's an unkit? It's a standard Econoram board that has all sockets and bypass caps pre-soldered in place. To complete assembly, the user simply solders in a few other parts, and inserts all ICs into their sockets. The result: A one-evening project that saves money while offering true CompuPro/Econoram quality for those on a budget. Static technology used throughout; all boards except Econoram VI run with 4 MHz systems. Same 1 year limited warranty, same great specs as our regular boards.

Speaking of regular boards, we offer assembled/tested models and boards qualified under our high-reliability Certified System Component (CSC) program (200 hour burn-in, immediate replacement in event of failure within 1 year of invoice date). Refer to chart below for pricing.

Name	Storage	Buss	Configuration	Unkit	Assm	CSC
Econoram IIA	8K X 8	S-100	2-4K blocks	\$149	\$179	\$239
Econoram IV	16K X 8	S-100	1-16K	\$269	\$329	\$429
Econoram VI	12K X 8	H8	1-8K, 1-4K	\$200	\$270	n/a
Econoram VIIA-16	16K X 8	S-100	2-4K, 1-8K	\$279	\$339	\$439
Econoram VIIA-24	24K X 8	S-100	2-4K, 2-8K	\$398	\$485	\$605
Econoram IX-16	16K X 8	Dig Grp	2-4K, 1-8K	\$319	\$379	n/a
Econoram IX-32	32K X 8	Dig Grp	2-4K, 1-8K, 1-16K	\$559	\$639	n/a
Econoram X	32K X 8	S-100	2-8K, 1-16K	\$529	\$649	\$789
Econoram XI	32K X 8	SBC	2-8K, 1-16K	n/a	n/a	\$1050

BANK SELECT MEMORIES (for Alpha Micro Systems, Marinchip, etc.)

Econoram XII-16	16K X 8	S-100	2 ind. banks**	\$329	\$419	\$519
Econoram XII-24	24K X 8	S-100	2 ind. banks**	\$429	\$539	\$649
Econoram XIII	32K X 8	S-100	2 ind. banks**	\$559	\$699	\$849

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7400 TTL

SN7400N	16	SN7470N	29
SN7401N	16	SN7472N	29
SN7402N	18	SN7473N	35
SN7403N	18	SN7474N	35
SN7404N	18	SN7475N	40
SN7405N	20	SN7480N	50
SN7406N	29	SN7482N	99
SN7407N	29	SN7483N	59
SN7408N	29	SN7485N	79
SN7409N	20	SN7486N	35
SN7410N	18	SN7489N	175
SN7411N	25	SN7490N	89
SN7412N	25	SN7491N	59
SN7413N	40	SN7492N	43
SN7414N	70	SN7493N	43
SN7415N	25	SN7494N	65
SN7416N	25	SN7495N	65
SN7420N	20	SN7496N	65
SN7421N	29	SN7497N	300
SN7422N	39	SN7498N	89
SN7423N	39	SN7499N	35
SN7425N	29	SN7499N	99
SN7426N	29	SN7415N	195
SN7427N	25	SN7419N	35
SN7429N	39	SN7419N	125
SN7430N	20	SN7419N	49
SN7432N	25	SN7419N	195
SN7437N	25	SN7419N	49
SN7438N	25	SN7419N	75
SN7439N	25	SN7419N	35
SN7440N	20	SN7419N	99
SN7441N	89	SN7419N	149
SN7442N	49	SN7419N	29
SN7443N	75	SN7414N	295
SN7444N	75	SN7414N	295
SN7445N	75	SN7417N	195
SN7446N	69	SN7418N	125
SN7447N	59	SN7415N	89
SN7448N	79	SN7415N	59
SN7449N	79	SN7415N	99
SN7450N	99	SN7415N	295
SN7451N	20	SN7415N	59
SN7453N	20	SN7415N	99
SN7454N	20	SN7415N	79
SN7459N	25	SN7415N	79
SN7460N	20	SN7415N	65

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JE600 HEXADECIMAL ENCODER KIT



FEATURES
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• LED readout to verify entries
• Easy interfacing with standard 16 pin IC connector
• Only -5VDC required for operations

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200+ dis.
2nd dia.

XCS55R	red	5/31
XCS55G	green	4/51
XCS55Y	yellow	4/51
XCS55C	clear	4/51
XC209R	red	5/31
XC209G	green	4/51
XC209Y	yellow	4/51
XC209C	clear	4/51
XC22R	red	5/31
XC22G	green	4/51
XC22Y	yellow	4/51
XC22C	clear	4/51
MV10R	red	4/51
MV50R	red	6/51
INFRA-RED LED	1/4" x 1/16" flat	XC111G green XC111Y yellow XC111C clear

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JE300 \$39.95

TIMEX T1001 LIQUID CRYSTAL DISPLAY CLASS II FIELD EFFECT



4 DIGIT - 5" CHARACTERS
THREE ENUNCIATORS
2.00" x 1.20" PACKAGE
INCLUDES CONNECTOR

T1001-Transmissive \$7.95
T1001A-Reflective \$8.25

DISPLAY LEDS

TYPE POLARITY HT PRICE TYPE POLARITY HT PRICE

MAN 1	Common Anode-red	270	MAN 6730	Common Anode-red +	560	99
CD4001	23	CD4001	23	CD4001	5 x 7	99
CD4002	23	CD4028	89	CD4072	49	19
CD4006	1 19	CD4029	1 19	CD4076	49	19
CD4007	25	CD4030	49	CD4061	23	19
CD4009	49	CD4032	99	CD4062	23	79
CD4010	49	CD4040	1 19	CD4093	99	19
CD4011	23	CD4041	1 25	CD4096	24	79
CD4012	25	CD4042	99	MC14409	14 95	80
CD4013	39	CD4043	89	MC14410	14 95	84
CD4014	39	CD4044	89	MC14411	14 95	80
CD4015	19	CD4046	4 95	MC14412	14 95	80
CD4016	49	CD4047	2 50	MC14433	14 95	80
CD4017	1 19	CD4048	1 38	MC14506	19 95	95
CD4018	99	CD4049	49	MC14507	19 95	95
CD4019	49	CD4050	49	MC14562	14 95	71
CD4020	1 19	CD4051	1 19	MC14583	3 50	71
CD4021	1 19	CD4053	1 19	MC14601	1 19	71
CD4022	1 19	CD4054	2 95	MC14611	1 19	71
CD4023	23	CD4055	2 95	MC14613	1 19	71
CD4024	23	CD4056	2 95	MC14614	1 19	71
CD4025	23	CD4057	2 95	MC14615	1 19	71
CD4026	23	CD4058	2 95	MC14616	1 19	71
CD4027	69	CD4060	45	MC14617	1 19	71

74C00

74C00	39	74C02	24
74C04	39	74C05	24
74C08	29	74C09	24
74C10	39	74C19	24
74C14	1 95	74C19	24
74C20	39	74C22	24
74C30	39	74C51	2 95
74C42	1 95	74C52	1 95
74C48	2 95	74C56	2 95
74C49	2 95	74C57	2 95
74C50	2 95	74C58	2 95
74C51	2 95	74C59	2 95
74C52	2 95	74C60	2 95
74C53	2 95	74C61	2 95
74C54	2 95	74C62	2 95
74C55	2 95	74C63	2 95
74C56	2 95	74C64	2 95
74C57	2 95	74C65	2 95
74C58	2 95	74C66	2 95
74C59	2 95	74C67	2 95
74C60	2 95	74C68	2 95
74C61	2 95	74C69	2 95
74C62	2 95	74C70	2 95
74C63	2 95	74C71	2 95
74C64	2 95	74C72	2 95
74C65	2 95	74C73	2 95
74C66	2 95	74C74	2 95
74C67	2 95	74C75	2 95
74C68	2 95	74C76	2 95
74C69	2 95	74C77	2 95
74C70	2 95	74C78	2 95
74C71	2 95	74C79	2 95
74C72	2 95	74C80	2 95
74C73	2 95	74C81	2 95
74C74	2 95	74C82	2 95
74C75	2 95	74C83	2 95
74C76	2 95	74C84	2 95
74C77	2 95	74C85	2 95
74C78	2 95	74C86	2 95
74C79	2 95	74C87	2 95
74C80	2 95	74C88	2 95
74C81	2 95	74C89	2 95
74C82	2 95	74C90	2 95
74C83	2 95	74C91	2 95
74C84	2 95	74C92	2 95
74C85	2 95	74C93	2 95
74C86	2 95	74C94	2 95
74C87	2 95	74C95	2 95
74C88	2 95	74C96	2 95
74C89	2 95	74C97	2 95
74C90	2 95	74C98	2 95
74C91	2 95	74C99	2 95
74C92	2 95	74C100	2 95
74C93	2 95	74C101	2 95
74C94	2 95	74C102	2 95
74C95	2 95	74C103	2 95
74C96	2 95	74C104	2 95
74C97	2 95	74C105	2 95
74C98	2 95	74C106	2 95
74C99	2 95	74C107	2 95
74C100	2 95	74C108	2 95
74C101	2 95	74C109	2 95
74C102	2 95	74C110	2 95
74C103	2 95	74C111	2 95
74C104	2 95	74C112	2 95
74C105	2 95	74C113	2 95
74C106	2 95	74C114	2 95
74C107	2 95	74C115	2 95
74C108	2 95	74C116	2 95
74C109	2 95	74C117	2 95
74C110	2 95	74C118	2 95
74C111	2 95	74C119	2 95
74C112	2 95	74C120	2 95
74C113	2 95	74C121	2 95
74C114	2 95	74C122	2 95
74C115	2 95	74C123	2 95
74C116	2 95	74C124	2 95
74C117	2 95	74C125	2 95
74C118	2 95	74C126	2 95
74C119	2 95	74C127	2 95
74C120	2 95	74C128	2 95
74C121	2 95	74C129	2 95
74C122	2 95	74C130	2 95
74C123	2 95	74C131	2 95
74C124	2 95	74C132	2 95
74C125	2 95	74C133	2 95
74C126	2 95	74C134	2 95
74C127	2 95	74C135	2 95
74C128	2 95	74C136	2 95
74C129	2 95	74C137	2 95
74C130	2 95	74C138	2 95
74C131	2 95	74C139	2 95
74C132	2 95	74C140	2 95
74C133	2 95	74C141	2 95
74C134	2 95	74C142	2 95
74C135	2 95	74C143	2 95
74C136	2 95	74C144	2 95
74C137	2 95	74C145	2 95
74C138	2 95	74C146	2 95
74C139	2 95	74C147	2 95
74C140	2 95	74C148	2 95
74C141	2 95	74C149	2 95
74C142	2 95	74C150	2 95
74C143	2 95	74C151	2 95
74C144	2 95	74C152	2 95
74C145	2 95	74C153	2 95
74C146	2 95	74C154	2 95
74C147	2 95	74C155	2 95
74C148	2 95	74C156	2 95
74C149	2 95	74C157	2 95
74C150	2 95	74C158	2 95
74C151	2 95	74C159	2 95
74C152	2 95	74C160	2 95
74C153	2 95	74C161	2 95
74C154	2 95	74C162	2 95
74C155	2 95	74C163	2 95
74C156	2 95	74C164	2 95
74C157	2 95	74C165	2 95
74C158	2 95	74C166	2 95
74C159	2 95	74C167	2 95
74C160	2 95	74C168	2 95
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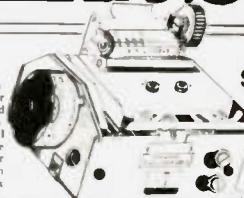
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1-12VDC SPDT RELAY, 180 ohm coil, 25 mA, 12x1x1" (#5937)	1.29	2 for 1.30
15-VOLUME CONTROLS, ass't, audio, and switch too! (#592)	1.29	30 for 1.30
75-PREFORMED DISC CAPS, handy assortment of values, marked, (#1181)	1.29	150 for 1.30
15-AXIAL ELECTROS, ass'td values and capacitance, (#5901)	1.29	30 for 1.30
100-GLASS ZENERS, 400mw, untested, better than 50% yield, (#5889)	1.29	200 for 1.30
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30-TWO WATT RESISTORS, carbon-films, carbons, some 5-wkers, (#456)	1.29	60 for 1.30
60-TERMINAL STRIPS, ass'td solder and screw types, 2 lugs & up, (#334)	1.29	120 for 1.30
1-WATCH GUTS, LED, who knows how good, micro-digital bonanza, (#5115)	1.29	2 for 1.30
10-1000V 1A RECTIFIERS, 1N4007, space case, axial leads, (#5926)	1.29	20 for 1.30
5-MULTI-DIGIT LED READOUTS, bubble magnifier, 2 to 6 digits, (#3624)	1.29	10 for 1.30
15-POWER TAB TRANSISTORS, NPN, plastic, TO-220, (#5898)	1.29	30 for 1.30
6-PRECISION TRIM POT'S, ass'td singles and multi-turns, (#3389)	1.29	12 for 1.30
50-1N4000 RECTIFIERS, epoxy, axial leads, untested, (#2594)	1.29	100 for 1.30
30pc-HEAT SHRINK, Thermo-fit, useful ass't of sizes, shrinks 50%, (#5248)	1.29	60 for 1.30
15-SLIDE SWITCHES, SPST, SPDT, etc, all shapes and sizes, (#5927)	1.29	30 for 1.30
25-DTL's, 100% prime, ass'td flip flops, etc., marked, (#3709)	1.29	50 for 1.30
25-TTL's, 100% prime, ass'td flip flops, etc., marked, (#5873)	1.29	50 for 1.30
6-HOBBY OPTO COUPLERS, 1500 volt isolation, U-test, (#2629A)	1.29	12 for 1.30
8-TAPE RECORDER EARPHONE, for radios, recorders, 8 ohms, (#294b)	1.29	16 for 1.30
8-MICRO MINI JACKS, in a block, for 2mm sub-mini plugs, (#1437)	1.29	16 for 1.30
5-PL-55 PHONE JACKS, standard bushing, for home, communications, (#5868)	1.29	10 for 1.30
15-RED DEVIL CAPACITORS, handy assort of pop. values, axial leads, (#3823)	1.29	30 for 1.30
60-MICAS ass't, sizes-n-shapes incl., "silvers", too! (#373)	1.29	120 for 1.30
12-TRANSISTOR SOCKETS, for npn and pnp types, (#5944)	1.29	24 for 1.30
12-LED DRIVER IC's, similar to 75491, (#5890)	1.29	24 for 1.30
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6-MOTORS MOTORS", small, high speed, ass'td slizes, 3-6VDC, (#2551)	1.29	12 for 1.30
2-HEAVY DUTY LINE CORDS, 8 ft., 2 cond., 18 gauge, black insulation, (#5803)	1.29	4 for 1.30
12-LED SOCKETS, "snap-in", minis, for LEDs, and translators too! (#5745)	1.29	24 for 1.30
15-MV-54 MICRO-MINI LED'S, flat top-style, diffused red lens, (#1789)	1.29	30 for 1.30
30-WIRE NUTS, twist-on's, for 20-24 gauge wire, (#3724)	1.29	60 for 1.30
2-ALUM. HEAT SINKS, 2 3/4" x 3", for power tabs, innumerable uses, (#5338)	1.29	4 for 1.30
1-MICRO MINI METER, 1/2" dia, 0-1 base movement, (#5858)	1.29	2 for 1.30
2-DOUBLE-SIDED PC BOARDS, 3" x 12", high quality G-10 glass, (#5694)	1.29	4 for 1.30
60-TUBULAR CAPACITORS, ass'td 100mfd to 1mf, up to 600 VVDC, (#354219)	1.29	120 for 1.30
6-MICRO MINI REED SWITCHES, 1" long, for alarms, relays, etc., (#1258)	1.29	12 for 1.30
10-TANTELUM ELECTROS, ass'td mini, axial, hermetically sealed, (#5848)	1.29	20 for 1.30
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1-TO-3 HEAT SINK, heavy duty aluminum, pre-punched for TO-3, (#4083)	1.29	2 for 1.30
75-MOLEX SOCKETS, "on-a-strip", for multiple pin dips, (#1609)	1.29	150 for 1.30
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1-UHF TUNER, 3 gang, uhf TV until, 300 ohm, pos. direct drive, (#2927)	1.29	2 for 1.30
15-HUMBUCKER CONTROLS, ass'td values, manufacturers dump, (#3807)	1.29	30 for 1.30
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POTTER & BRUM-FIELD 12VDC RELAYS

SPDT 25 Amps \$5.95
2PDT 3 Amps \$1.99

KILOVAC H-8/54

VACUUM RELAY

26 volt coil \$29.95 each

E. F. JOHNSON TUBE SOCKETS

for 3-400Z and
3-5000Z tubes \$29.95 per pair

TOROIDAL CORES

T-37-6 6/1.00
25/4.00, 50/6.00
100/10.00

4CX250B/R SOCKETS AND CHIMNEYS NEW

\$14.95 per set (1 socket, 1
chimney)

B&W COILS

1206T \$3.99
2006T \$7.99

FAIRCHILD REGULATOR 78H05KC

\$6.99 each

TUBES

6146B \$6.50

MINIMUM ORDER \$5.00
Minimum Shipping \$1. Insurance 35¢
per \$100. COD charges 85¢ to street address only!
We prefer street address as we ship UPS and P.O. Box it's take up to
50% longer to deliver. We accept VISA
or Mastercharge. Please list complete
card number and expiration date. Allow
10% extra for shipping of heavy items.
We reserve the right to change prices
without notice. All items listed are sub-
ject to prior sale. Some items listed are
in small quantities.

BULLET ELECTRONICS

P.O. BOX 401244E
GARLAND, TX 75040
(214) 278-3553

WE APOLOGIZE!

JUST WHEN IT LOOKED LIKE ALL THE HOOPLA ABOUT OUR LOW COST HIGH CURRENT REGULATED PS-14 POWER SUPPLY KIT WAS FADING AWAY AND OUR COMPETITORS COULD BREATHE A LITTLE EASIER, WE GO AND DO SOMETHING LIKE THIS!

PS-15 POWER SUPPLY KIT 12V 30 AMPS

SPECS:

CONTINUOUS CURRENT	30 AMPS
REGULATION	200 MV LOAD AND LINE
RISSLE	LESS THAN 300MV @ 30A
CURRENT LIMIT	ADJUSTABLE FOLDBACK TYPE
VOLTAGE OUTPUT	ADJUSTABLE 11.5 TO 14 VDC
SAFETY FEATURES	SHORT CIRCUIT PROTECTION; THERMAL SHUTDOWN; RF BY-PASSED

PRICE DOES NOT INCLUDE CHASSIS, METERS OR JACKS

PS-14 Power Supply Kit.

Same features as above but 20 AMP max current. SHIPPING PAID.

45.00

Regulator Card &
Components: **14.95**

YOU GET:

- HUGE 40 AMP TRANSFORMER
- 3 LARGE HEATSINKS
- DRILLED & PLATED PC BOARD
- LARGE COMPUTER GRADE CAP
- ALL ELECTRONIC PARTS
- FUSES, FUSEHOLDERS, LINE CORD WIRE, MICA INSULATORS
- COMPLETE STEP-BY-STEP INSTRUCTIONS
- COMPLETE WARRANTY AND REPAIR SERVICE
- WE PAY SHIPPING!

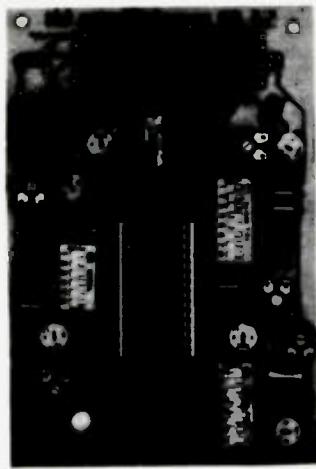
59.95

*Canadian Orders add \$10.00 for Shipping and Insurance.

OVP-2 OVERVOLTAGE PROTECTOR

Protect your expensive gear from overvoltage from transients or power supply malfunctions. Use with any fused power supply. Compatable with PS-14 or PS-15. Adjustable 10-20 VDC

SE-01 SOUND EFFECTS KIT



- ALL PARTS
- COMPLETE INSTRUCTIONS
- QUALITY PLATED & DRILLED BOARD
- IC SPECS INCLUDED
- RUNS ON ONE 9-VOLT BATTERY
- SMALL 3 1/4" x 5"
- **\$16.95**
LESS SPEAKER & BATTERY
- 76477 CHIP IS INCLUDED. EXTRA CHIPS \$2.95 EACH

OTHERS WILL SELL YOU THE CHIP. BUT ONLY BULLET CAN FURNISH YOU A COMPLETE KIT OF ALL THE PARTS YOU NEED TO PUT THE T.I. 76477 SOUND CHIP THROUGH ITS PACES! WE INCLUDE SPECS AND PROGRAMMING CHARTS. YOU WILL BE AMAZED AT THE THOUSANDS OF SOUND EFFECTS YOU CAN PRODUCE WITH OUR KIT! MAKE PHASOR GUN, STEAM TRAIN, GUNSHOT AND OTHER SOUNDS. BOARD HAS AUXILIARY PULSE GENERATOR, COMPARATOR AND MULTIPLEX OSC FOR EVEN MORE VERSATILITY.

MATCHED METER SET

A quality pair of American Made 3 1/2" rectangular panel meters. 5% Accuracy. Modern styling.
SET 0-15VDC. 0-30ADC

12.95 set

7 WATT AUDIO AMP KIT

SMALL, SINGLE HYBRID IC AND COMPONENTS FIT ON A 2" x 3" PC BOARD (INCLUDED). RUNS ON 12 VDC. GREAT FOR ANY PROJECT THAT NEEDS AN INEXPENSIVE AMP. LESS THAN 3% THD @ 5 WATTS. COMPATIBLE WITH SE-01 SOUND KIT.

\$5.95

6 DIGIT AUTO/VAN CLOCK

- LARGE 1/2" CHARACTERS (LED)
- QUARTZ XTAL TIMEBASE
- ALARM & SNOOZE OPTIONS
- NOISE FILTERING
- EASY TO ASSEMBLE
- 4 1/2" x 3" x 1 1/2"
- DRILLED & PLATED PC BOARDS

COMPLETE KIT

12 VDC

\$16.95

NEW!

ULTRASONIC RELAY KIT

INVISIBLE BEAM WORKS LIKE A PHOTO ELECTRIC EYE. USE UP TO 25 FT. APART. COMPLETE KIT. ALL PARTS & PC BOARDS.

\$21.50

SPECIAL PACKAGE:

PS-15, OVP-2
AND METERS

75.00

SHIPPING PAID

PARTS

301 DP AMP 8 LEAD CAN	3/1.00
723 VOLT REG. 10 LEAD CAN	.50
13741 FET INPUT 741 MINI DIP	3/1.10
30.000 @ 15V COMPUTER GRADE	2.10
2N4400 NPN GEN. PURPOSE	8/1.00
2N4402 PNP COMPLIMENT	8/1.00
2N6028 P.U.T. W/SPEC'S	.50
LM380 2W AUDIO IC W/SPEC'S	1.09
LM377 DUAL LM380 W/SPEC'S	2.50
*7215 VOLT REG. 1A 15V	.69
*725 LOW NOISE OP AMP	.99
IL-1 OPTO ISOLATOR MINI DIP	.60
*MEM 631 DUAL GATE MOSFET.	
DIODE PROTECTED. SIMILAR	
TO 40673	.50
MV1624 VARICAP DIODE 10 PFD	.49
IN4003 1A 200V DIODE	15/1.00
TIP30 TAB PNP POWER	3/1.00
*MC1351P FM IF. DISC IC	.50

*INDICATES ITEM IS "HOUSE NUMBERED"

LED'S

JUMBO GREEN	4/.89
JUMBO RED	5/.89
MEDIUM RED (1/4")	.15
MEDIUM GRN OR YELLOW	.16

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- SEND CHECK, M.O.,

M.C. OR VISA

- ADD 5% FOR

SHIPPING

- CATALOG FREE UPON REQUEST

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- FOREIGN ADD 10%

DIODES/ZENERS			
QTY.	1N914	100v	10mA
	1N4005	600v	1A
	1N4007	1000v	1A
	1N4148	75v	10mA
	1N4733	5.1v	1 W Zenner
	1N4749	24v	1W
	1N753A	6.2v	500 mW Zener
	1N758A	10v	"
	1N759A	12v	"
	1N5243	13v	"
	1N5244B	14v	"
	1N5245B	15v	"
	1N5349	12v	3W

SOCKETS/BRIDGES			
QTY.	8-pin	pcb	.16 ww .35
	14-pin	pcb	.20 ww .40
	16-pin	pcb	.25 ww .45
	18-pin	pcb	.30 ww .95
	20-pin	pcb	.35 ww 1.05
	22-pin	pcb	.40 ww 1.15
	24-pin	pcb	.45 ww 1.25
	28-pin	pcb	.50 ww 1.35
	40-pin	pcb	.55 ww 1.45
	Molex pins	.01 To-3 Sockets	.35
	2 Amp Bridge	100-prv	.95
	25 Amp Bridge	200-prv	1.50

TRANSISTORS, LEDs, etc.			
QTY.	2N2222M	(2N2222 Plastic .10)	.15
	2N2222A		.19
	2N2907A	PNP	.19
	2N3906	PNP (Plastic)	.19
	2N3904	NPN (Plastic)	.19
	2N3054	NPN	.55
	2N3055	NPN 15A 60v	.60
	T1P125	PNP Darlington	1.95
	LED Green, Red, Clear, Yellow	.19	
	D.L.747	7 seg 5/8" High com-anode	1.95
	MAN72	7 seg com-anode (Red)	1.25
	MAN3610	7 seg com-anode (Orange)	1.25
	MAN82A	7 seg com-anode (Yellow)	1.25
	MAN74	7 seg com-cathode (Red)	1.50
	FND359	7 seg com-cathode (Red)	1.25

9000 SERIES			
QTY.	9301	.85	9322 .65
	9309	.50	9601 .30
			9602 .45

C MOS			
QTY.	QTY.	QTY.	QTY.
4000	.15	4017	.75
4001	.20	4018	.75
4002	.25	4019	.35
4004	3.95	4020	.85
4006	.95	4021	.75
4007	.25	4022	.75
4008	.75	4023	.25
4009	.35	4024	.75
4010	.35	4025	.25
4011	.30	4026	1.95
4012	.25	4027	.35
4013	.40	4028	.75
4014	.75	4029	1.15
4015	.75	4030	.30
4016	.35	4033	1.50
			4066 .75
			74C151 2.50

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CHARGE CARD # AE Visa BA MC

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SPECIAL DISCOUNTS

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WILL CALL _____

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NET 10th OF THE MONTH _____

PO # _____

Total Order \$35-\$99 Deduct 10%

ALL ORDERS SHIPPED PREPAID - NO MINIMUM - COD ORDERS ACCEPTED - ALL ORDERS SHIPPED SAME DAY

OPEN ACCOUNTS INVITED - California Residents add 6% Sales Tax. PRICES SUBJECT TO CHANGE WITHOUT NOTICE.

\$100-\$300 15%

\$301-\$1000 20%

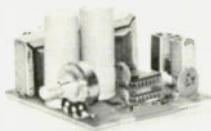
24 Hour Phone Service — We accept American Express / Visa / BankAmericard / Master Charge

- T T L -			
QTY.	7400	.20	7492 .45
	7401	.20	7493 .35
	7402	.20	7494 .75
	7403	.20	7495 .60
	7404	.20	7496 .80
	7405	.35	74100 1.15
	7406	.25	74107 .35
	7407	.55	74121 .35
	7408	.20	74122 .55
	7409	.25	74123 .55
	7410	.20	74125 .45
	7411	.25	74126 .45
	7412	.25	74132 .75
	7413	.45	74141 .90
	7414	.75	74150 .85
	7416	.25	74151 .95
	7417	.40	74153 .95
	7420	.25	74154 1.15
	7426	.25	74156 .70
	7427	.25	74157 .65
	7430	.20	74161/9316 .75
	7432	.30	74163 .85
	7437	.20	74164 .75
	7438	.30	74165 1.10
	7440	.20	74166 1.75
	7441	1.15	74175 .90
	7442	.55	74176 .95
	7443	.45	74177 1.10
	7444	.45	74180 .95
	7445	.75	74181 2.25
	7446	.70	74182 .75
	7447	.70	74190 1.25
	7448	.50	74191 1.25
	7450	.25	74192 .75
	7451	.25	74193 .85
	7453	.20	74194 .95
	7454	.25	74195 .95
	7460	.40	74196 .95
	7470	.45	74197 .95
	7472	.40	74198 1.45
	7473	.25	74221 1.50
	7474	.30	74298 1.50
	7475	.35	74367 1.35
	7476	.40	75491 .65
	7480	.75	75492 .65
	7481	.85	74H00 .20
	7482	.95	74H01 .30
	7483	.95	74H04 .30
	7485	.75	74H05 .25
	7486	.55	74H08 .35
	7489	1.05	74H10 .35
	7490	.55	74H11 .25
	7491	.70	74H15 .45

I ² L, LINEARS, REGULATORS, ETC.			
QTY.	QTY.	QTY.	QTY.
MCT2	.95	LM320K24	1.65
8038	3.95	LM320T5	1.65
LM201	.75	LM320T12	1.65
LM301	.45	LM320T15	1.65
LM308	.65	LM323K	5.95
LM309H	.85	LM324	1.25
LM309 (340K-5)	1.50	LM339	.75
LM310	.85	7805 (340T5)	1.15
LM311 (8-14 Pin)	.75	LM340T12	.95
LM318	1.50	LM340T15	.95
LM320H6	.79	LM340T18	.95
LM320H15	.79	LM340T24	.95
LM320H24	.79	LM340K12	1.25
7905 (LM320K5)	1.65	LM340K15	1.25
LM320K12	1.65	LM340K18	1.25
LM320K15	1.65	LM340K24	1.25
			LM1458 .65
			LM3900 .95
			LM75451 .65
			NE555 .45
			NE556 .85
			NE565 1.15
			NE566 1.25
			NE567 .95
			TA7205 6.95
			76477 2.95
			95H90 9.95

VARIABLE POWER SUPPLY KIT \$11⁹⁵

- Continuously Variable from 2V to over 15V
- Short-Circuit Proof
- Typical Regulation of 0.1%
- Electronic Current Limiting at 300mA
- Very Low Output Ripple
- Fiberglass PC Board Mounts All Components
- Assemble in about One Hour
- Makes a Great Bench or Lab Power Supply
- Includes All Components except Case and Meters



ADD \$1.25 FOR POSTAGE/HANDLING

FREE

IC or FET's WITH
\$5 & \$10 ORDERS.
DATA SHEETS
WITH MANY ITEMS.

SPECIALS - THIS MONTH ONLY

IN34A Germanium Diode 88V 10mA	10/51	LM308H Low Bias Current Op Amp-Super 700	\$0.84
IN270 Germanium Diode 88V 200mA	4/51	LM309K 5 Volt Regulator	TD-3 .84
IN914 Silicon Diode 100V 10mA	26/51	LM317K Adjustable Voltage Regulator	2-37V 3.50
IN8263 Hot Carrier Diode (HP2800, etc.)	\$1.00	LM380K 2 Watt Audio Power Amplifier	DIP .84
F7 Power Varistor 1-2W D=432MHz (Specs & Circuits included with F7)	\$2.00	NE560A Phase Locked Loop	DIP .84
DIODE GRAB BAG-Mixed zeners, rectifiers, etc.	56/51	LMT23CN Precision Voltage Regulator	DIP 3/\$1
2N706 NPN High Speed Switch 75ns	4/51	LM747 Dual 741 Compensated Op Amp	DIP 2/\$1
2N510 UHF Transistor-Osc/Amp up to 1 GHz	4/51	2102 1024-Hk State RAM (1024 x 1)	DIP \$1.75
2N2609 P-Channel FET Amplifier 2500mhz	\$1.00	27400E FET-Input Op Amp-like NE526/UA740	1.95
2N2622 NPN Dual Transistor 3mV Match ±25	2.95	CA3018A 4-Transistor Array/Darlington	.99
2N3904 NPN Op/Amp/Switch ±100 40V 200mA	5/51	CA3028A RF/IF Amplifier DC to 120MHz	1.45
2N4122 PNP RF Amplifier & Switch	3/51	CA3075 FM IF Op/Amp/Limiter/Detector	DIP 1.45
2N4886E N-Channel Audio FET Super Low-Noise	2/51	RC4558 Dual High Gain Op Amp	mDIP 3/51
2N4886 150 Volt PNP Transistor for Keyer	2/51	NS555V Precision Fast Op Amp	mDIP 2/51
E112 N-Channel FET VHF RF Amp	3/51	NS555BV Dual Hl Gain Op Amp Comp.	mDIP 3/51
TS74 N-Channel FET High-Speed Switch 40ns	3/51	8038 Function Generator/VCO with circuits	\$3.75

LP-10 LOGIC PROBE bit-TTL, CMOS, etc.

Machined case included -½ hr. assembly
\$7.85
SEND FOR ADVA'S NEW 1979 CATALOG
NEARLY 1000 SEMICONDUCTORS, KITS, CAPACITORS, ETC.-SEND 25¢ STAMP.

OTHER ADVA KITS:

LOGIC PROBE KIT-Use with CMOS, TTL, DTL, RTL, HTL, HNIL and most MOS IC's. Built-in protection against polarity reversal and overvoltage. Drives only a few mA from circuit under test. Dual LED readout. Complete kit includes case and clip leads. ONLY \$7.85

FIXED REGULATED POWER SUPPLY KITS-Short-circuit proof with thermal current limiting. Compact size and typical regulation of 0.5% make these ideal for most electronic projects. Available for 5V @ 500mA, 6V @ 500mA, 9V @ 500mA, 12V @ 400mA, 15V @ 300mA. Specify voltage when ordering.

These easy-to-assemble kits include all components, complete detailed instructions and plated fiberglass PCB boards. Power supply kits do not include case or meters. Add \$1.25 per kit for postage and handling.

1 MAIL NOW! FREE DATA SHEETS supplied with many items from this ad. FREE ON REQUEST-741 Op Amp with every order of \$5 or more--749 Dual Op Amp or two 100 FET's with every order of \$10 or more. Postage paid on orders prior to 12/31/78. One free item per order ONLY TODAY--All items subject to prior sale and prices subject to change without notice. All items are new surplus parts--100% functionally tested.

WRITE FOR FREE CATALOG #79 offering over 700 semiconductors carried in stock. Send 25¢ stamp.

TERMS: Send check or money order (U.S. funds) with order. Add 5% postage for U.S., Canada and Mexico. \$1.00 handling charge on orders under \$10 Calif. residents add 6% sales tax. Foreign orders add 10% postage COD orders add \$1.00 service charge

MORE SPECIALS:

RC4195DN • 15V @ 50mA VOLTAGE REGULATOR IC. Very easy to use. Makes a nice Highly Regulated 15V Supply for OP AMP's, etc. Requires only unregulated DC (18-30V) and 2 bypass capacitors. With Data Sheet and Schematics. 8-pin mDIP \$1.25

RC4136 Quad 741 Low-Noise Op Amp mDIP \$0.95

LM1304 FM Multiplex Stereo Demodulator DIP 0.99

LM2111 FM IF Subsystem (IF Amp, Det., Limiter) DIP 0.99

1N6263 Hot Carrier Diode 0.4V @ 1mA 0.1mA DO-35 1.00

ZENERS-Specify Voltage 3.3, 3.9, 4.3, 5.1, 6.8, 8.2 400mW \$1.00

9.1, 10, 12, 15, 16, 18, 20, 22, 24, 27, or 33V (*10%) 1 Watt \$3.00

• MONEY-BACK GUARANTEE

• ALL TESTED AND GUARANTEED

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BOX 4181 GB, WOODSIDE, CA 94062
Tel. (415) 328-1500

*SUPER SPECIALS:

IN34 Germanium Diode	10/51	FSA2501M Diode Array	2/\$1
IN914 100V/10mA Diode	20/51	MPF102 200MHz RF Amp	3/\$1
IN4001 50V/1A Rectifier	15/51	40673 MOSFET RF Amp	\$1.75
IN4154 30V IN914	25/51	LM324 Quad 741 Op Amp	.94
BR1 50V ½A Bridge Rec	4/51	LM376 Pow Volt Reg mDIP	.56
2N2222 NPN Transistor	6/51	NE555 Timer mDIP	.38
2N2807 PNP Transistor	6/51	LM723 2-37V Reg OIP	3/\$1
2N3035 Power Xistor 10A	6/51	LM741 Comp Op Amp mOIP	6/51
2N3904 NPN Amp/Sw 100	6/51	LM1458 Dual 741 mDIP	3/\$1
2N3906 PNP Amp/Sw 100	6/51	CA3085 5 Trans Array DIP	.62
CP650 Power FET %Amp	55	RCA29 Pwr Xistor 1A 30W	.70

RF391 RF Power Amp Transistor 10-25W @ 3-30MHz TO-3
555X Timer 1s-1hr Different pinout from 555 (w/dt) \$5.00
RC4194T K Dual Tracking Regulator .02 to 30V @ 200mA TO-66
RC4195T K Dual Tracking Regulator 15V @ 100mA (TO-66) \$2.25
8038 Waveform Generator ~1A Wave With Circuits & Data \$3.75

NEW SPECIALS

LM318CN	High Speed Op Amp 50V/µs mDIP	\$0.94
LM318D	High Speed Op Amp 50V/µs DIP	.90
LM339N	Quad Comparator Single or Dual Supply	.79
LM380N-8	½W Audio Power Amplifier 8-22V	.90
NE567V	Tone Decoder (PLL) 0.01Hz to 500kHz	.99
XR567CP	Tone Decoder (PLL) 0.01Hz to 500 kHz	.99
LM723CN	Precision Voltage Regulator 2-37V DIP	3/51
LM747CN	Dual 741 Compensated Op Amp	2/\$1
SAD1024	Dual 512 Stage (1024) Audio Delay Line "Bucket Brigade" Appl. Data included	\$18.95
XR2206CP	Function Generator with applic. data	4.40
XR2242CP	Long-Range Precision Timer µs to days	1.50
1 YEAR TIMER Kit-2 XR2242's and Appl. Note	3.00	
LM2901N	Quad Comparator +5V or 2 to 36VDC	\$1.20
CA3018A	4-Transistor Array/Darlington	.99
CA3028A	RF/IF Amplifier DC to 120MHz	1.25
RC4558	Dual High Gain Op Amp mDIP	3/51
SPACER Kit - 4 ea of 5 assorted size spacers	\$1.00	
±15 VOLT Regulated Power Supply Kit-Easy assembly, 5mV regulation, 100mA, fully protected. Includes all components and hardware, no PCB or case-Add \$1.50 for shipping	\$13.95	

1N270	Germanium Diode 80V 200mA	4/\$1
1N823	Temp Comp Reference	\$0.60
1N914	6.2V ± 5% ± .005%/°C	
1N3044	100V Zener 1W-Better than an OB3	.75
1N3045	110V Zener 1W-Better than an OB2/OC3	.75
1N3071	200V 100mA Switching Diode 40ns	.30
2N2915	NPN Dual Transistor 3mV Match β100	\$1.95
2N3819M	N-Channel RF FET 100MHz Amp	.35
2N4020	PNP Dual Transistor 5mV Match β250	5.00
2N4445	N-Channel FET 5Ω Switch	3.50
2N5394E	Ultra-Low Noise J-FET Audio Amp	\$1.25
2N5912	Dual J-FET RF Diff Amp to 800MHz	2.90
2N6028	Programmable Unijunction Transistor	.45
2N6449	300 Volt N-Channel J-FET Amp/Sw	2.00
CP640	Broadband FET RF Amp 140dB	
E304	Dyn Range	\$4.50
MPSA12	N-Channel RF FET 3.8dB NF @ 400MHz	.50
TIS88	N-Channel Darlington Transistor β20,000	.75
RESISTOR Kit-150 pcs ½W, 20 most common values, individually packaged, 5 to 20 pcs each .	\$4.95	
PC BOARD Mounting Kit-8 ea spacer, screw, nut and washer-32 pcs total.	\$1.00	

SEND FOR ADVA'S NEW 1979 CATALOG

NEARLY 1000 SEMICONDUCTORS, KITS, CAPACITORS, ETC.-SEND 25¢ STAMP.

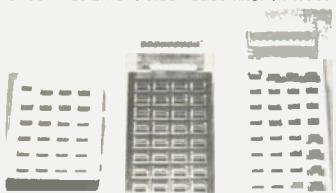
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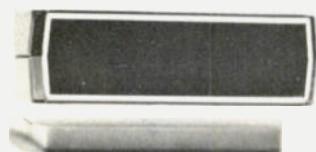
MOTOROLA 4 MHZ XTAL OSCILATOR 14 pin dip pkg., 5 VDC in, 4 MHz out. TTL compatible. \$5.95 ea.

NATIONAL SEMICONDUCTOR - 6 DIGIT CLOCK MODULE A 12-hr. clock featuring alarm and snooze alarm function. Comes complete with everything. All you add is a power cord and switches. Each module utilizes discreet components and is ready for you to mount in your enclosure. Display is 6 each seven segments, .33 inches high, clockchip is MM5375. Can easily be mounted in either of our instrument clock case kits. \$14.95.



CALCULATOR GUTS

Experimenter's delight. Each keyboard includes a monolithic calculator chip and display. These are rejects. It might be something simple to repair or it could be very nasty. At this price, who cares! 9 volt. \$1.50 ea. 3/\$3.00.



INSTRUMENT CLOCK CASE KIT

A real jewel for those small projects. Hinged top door allows you to hide your control area. O.D. 4 1/2" x 4" x 1 3/4". \$1.99 each.

DB-25 FEMALE CONNECTOR

A unique assembly originally adapting the RS232 Type Connector to flat cable. \$1.50 each or 2/\$2.50.



INSTRUMENT/CLOCK CASE KIT

Perfect for your opto projects. Solid aluminum construction with real walnut sides. O.D. 5-3/8" by 5-3/8" by 2". \$6.95 ea.



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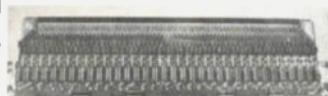
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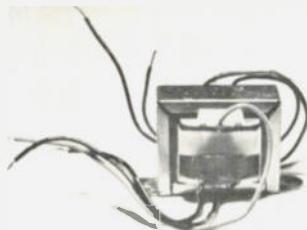
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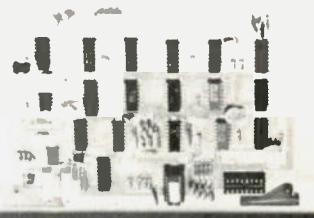
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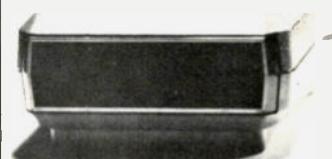
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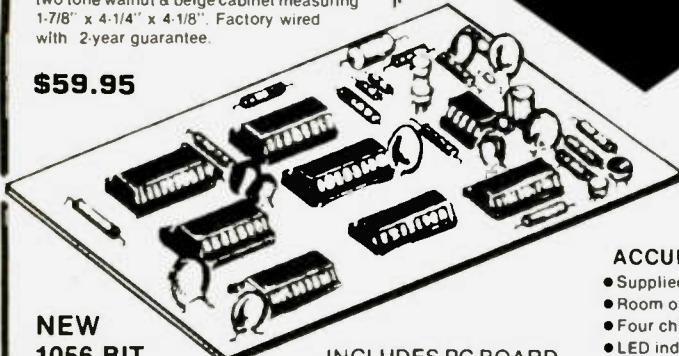


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SD1434	50W	5	450	500-6LFL	34.00	2N6084	40W	175 MHz	MT72	16.30
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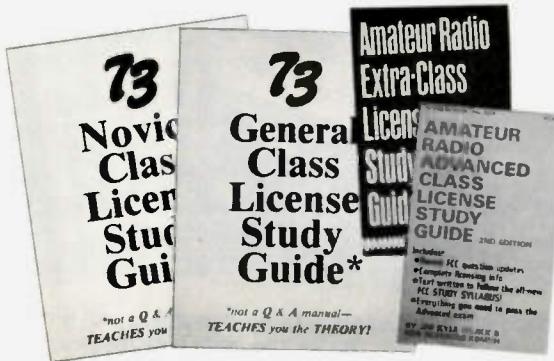
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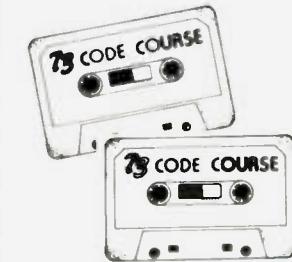
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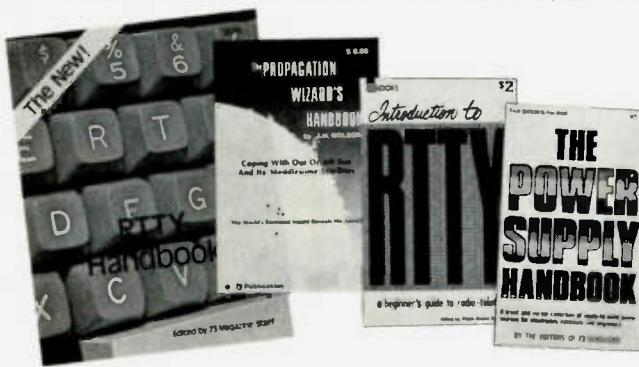
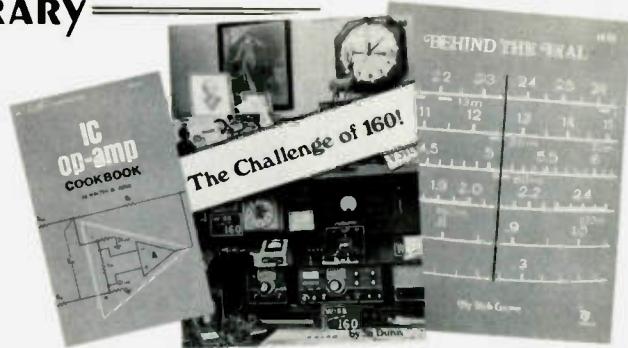
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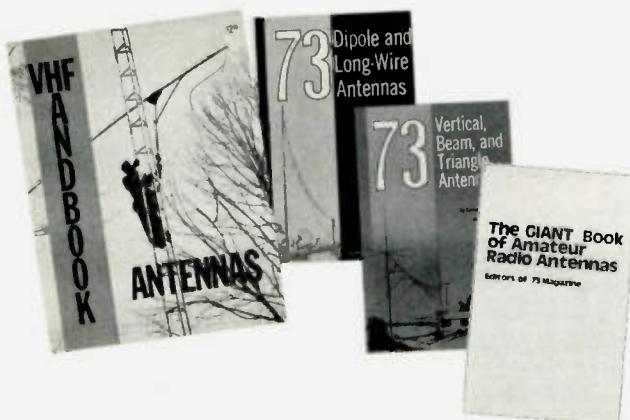


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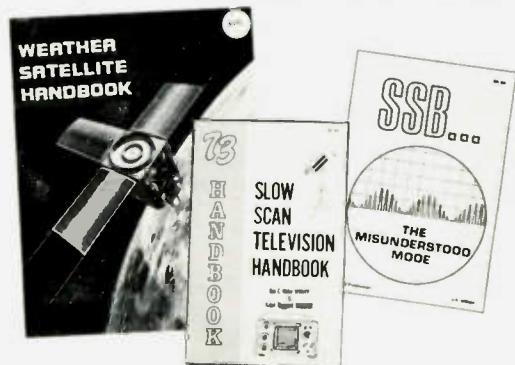
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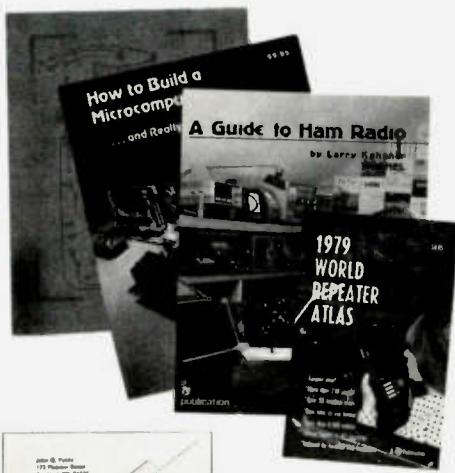
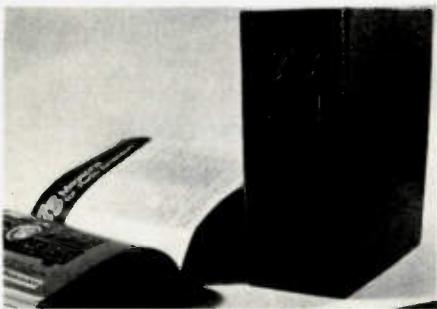
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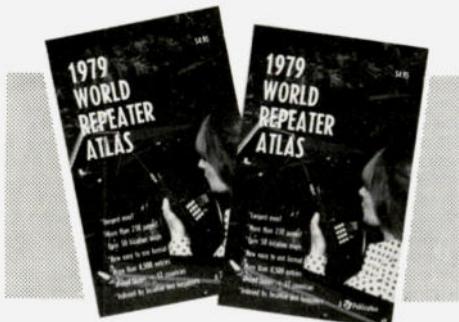
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A = Next higher frequency may also be useful

B = Difficult circuit this period

F = Fair

G = Good

P = Poor

SF = Chance of solar flares

September

SUN	MON	TUE	WED	THU	FRI	SAT
						1 G
2 G	3 F/SF	4 F/SF	5 F/SF	6 F/SF	7 F	8 G
9 G	10 G/SF	11 G/SF	12 P/SF	13 P/SF	14 P	15 G
16 G	17 F	18 G	19 G	20 G	21 G	22 G
23 G/F	24 F	25 G	26 G	27 G	28 G	29 G

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A46	ABC Communications.....	57	D23	Dovetron.....	166	M8	MHZ Electronics.....	170, 171	T18	Trac Electronics Corp.	77, 157
A1	Adirondack Radio Supply.....	125	D25	DSI Instruments.....	88, 89	M69	Micro Control Specialties.....	18	T3	Tufts Radio Elec.	164, 167
A24	Adva Electronics.....	186	E40	Electra.....	39	M95	Micro Management Systems.	18	U10	UDM Enterprises.	125
A60	AED Electronics.....	7	F1	Fair Radio Sales.....	125	M109	Mil Industries.	62	U13	Ultima Electronics, Ltd.	125
A92	AHF Antennas.....	105	F5	Flesher Corporation.....	95	N22	Non-Linear Systems.	23	U9	Unadilla/Reyco Division.	31
A107	AJT Enterprises.....	31	G12	Germantown Amateur Supply.....	24	N4	Nye Company.	24	U2	Unarco-Rohn.	77
A2	Aldeco.	188	G22	G.I.S.M.O.	127	O5	OK Machine & Tool.	24, 85	U8	United Products.	187
A40	Amateur Radio Supply of Nash- ville, Inc.	65	G4	Godbout Electronics.	59	O3	Optoelectronics, Inc.	13	V5	VHF Engineering.	41
A21	Amateur-Wholesale Electronics.....	188, 169	*	HAL Communications.	178	O12	Outdoor Outfitters.	27	V24	Vibroplex.	15, 109
A106	AMC Engineering.....	161	H24	Haf-Tronix.	21	*	Palomar Engineers.	121, 166	V75	Vineyard Amateur Radio.	125
*	Armsat.	99	H2	Ham Radio Center.....	107, 126	P20	para-graphs.	127	W18	Western Electronics.	135, 164
A6	Aptron Laboratories.	135	H16	Hamtronics, NY.	173	P41	P.C. Electronics.	84	W2	Wilson Electronics.	3
*	Associated Radio.	180	H8	Hamtronics, PA.	18	P44	Pickering Codemaster Company.....	107	W33	Wilson Systems, Inc.	91-94
A100	Auteck.	22	H5	Heath.	141, 158	P64	Rudy Plak.	125	X3	Xitex Corp.	99
B23	Barker-Williamson.	164	H3	Henry Radio.	CII	P2	Poly Pak.	181	Y1	Yaesu Electronics Corp.	CIII, 43, 75
B54	Barry Electronics.	130	H44	HFT, Inc.	125	Q3	Quest Electronics.	165	From 73.	79, 95, 130, 131-134, 189-194	
B42	Brodie Electronics Co.	117	*	ICOM.	19	R1	Radio Amateur Callbook.	146			
B8	Bullet Electronics.	184	I32	Instant Software.	86, 87	*	Radio World.	95			
C88	C & A Electronics Enterprises.....	31	I43	Insla-Pac.	24, 95	R8	Ramsey Electronics.	172			
C3	Clegg.	27, 164	I9	Integrated Circuits, Unltd.	185	R27	RF Power Labs, Inc.	145			
C21	Coakit.	7	I44	International Electronic Corp., Inc.	145	R35	Ross Distributing Corp.	127			
C106	Command Productions.	18	J1	Jameco Electronics.	179	S100	Scott Manufacturing.	130			
C58	Communications Center, NE.....	63, 90	J2	Jan Crystals.	135	S110	Selectone Corp.	47			
C5	Communications Elec.	18	J7	Jenson Tools & Alloys.	125	S63	Semiconductor Surplus.	183			
C115	Communications Electronics Specialties.	164, 166	*	Kantronics.	135, 146	S3	Sentry Manufacturing.	164			
C6	Communications Specialists.....	16, 17	K14	Kenwood.	CIV, 5, 139	S33	S-F Amateur Radio Serv.	7, 121			
C105	Communications & TV Unltd.....	161	K14	Key Electronics.	7	*	SoftSell Unlimited.	49			
D6	Peter W. Dahl Company.	55	L25	Kilobaud Microcomputing.	83	S81	Spectronics, Inc.	176, 177			
D35	Dayapro Electronics, Inc.	7	L9	The Logic Store.	95	S8	Spectrum Communications.	67			
D29	Dielectric Communications.	105	L17	Long's Electronics.	110-115	S10	SST Electronics.	169			
*	Digital Research Corporation.....	174	M48	Lunar Electronics.	24, 27	S18	Standard Communications.	84			
*	Digital Research Parts.	175	M35	Macrotronics, Inc.	105	S43	Surplus Electronics.	182			
			*	M36	Madison Electronics Supply.....	27, 62, 71, 126, 160	S44	Swan Electronics.	8, 9		
			M52	Maggiore Electronic Lab.	146	T64	Technical Clinic.	125			
				MFJ Enterprises.	25, 73	T52	Tele-Tow'r Mfg. Co.	121			
					T55	TET USA.	29				
					T34	Thomas Communications.	32, 33				
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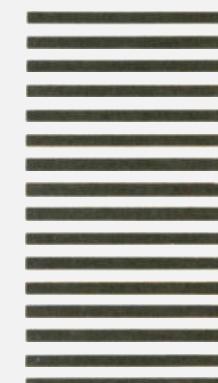
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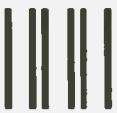
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Carrier Suppression: Better than 40 dB
Unwanted Sideband Suppression: Better than 40 dB @ 1000 Hz, 14 MHz
Spurious Radiation: Better than 40 dB below rated output
Third Order Distortion Products: Better than -31 dB
Transmitter Frequency Response: 300-2700 Hz (-6 dB)
Stability: Less than 300 Hz in first 30 minutes after 10 min. warmup; less than 100 Hz after 30 minutes over any 30 min. period
Negative Feedback: 6 dB @ 14 MHz
Antenna Output Impedance: 50-75 ohms, unbalanced

SPECIFICATIONS

GENERAL

Frequency Coverage: Amateur bands from 1.8-29.9 MHz, plus WWV/JJJ (receive only)
Operating Modes: LSB, USB, CW
Power Requirements: 100/110/117/200/220/234 volts AC, 50/60 Hz; 13.5 volts DC (with optional DC-DC converter)
Power Consumption: AC 117V: 75 VA receive (65 VA HEATER OFF) 285 VA transmit; DC 13.5V: 5.5 amps receive (1.1 amps HEATER OFF), 21 amps transmit
Size: 345 (W) x 157 (H) x 326 (D) mm
Weight: Approximately 15 kg.

COMPATIBLE WITH FT-901DM ACCESSORIES

RECEIVER

Sensitivity: 0.25 uV for S/N 10 dB
Selectivity: 2.4 KHz at 6 dB down, 4.0 KHz at 60 cB down (1.66 shape factor); Continuously variable between 300 and 2400 Hz (-6 dB); CW (with optional CW filter installed): 600 Hz at 6 dB down, 1.2 KHz at 60 dB down (2:1 shape factor)
Image Rejection: Better than 60 dB (160-15 meters); Better than 50 dB (10 meters)
IF Rejection: Better than 70 dB (160, 80, 20-10 m); Better than 60 dB (40 m)
Audio Output Impedance: 4-16 ohms
Audio Output Power: 3 watts @10% THD (into 4 ohms)



Price And Specifications Subject To Change Without Notice Or Obligation



YAESU ELECTRONICS CORP., 15954 Downey Ave., Paramount, CA 90723 • (213) 633-4007
 YAESU ELECTRONICS Eastern Service Ctr., 9812 Princeton-Glendale Rd., Cincinnati, OH 45246

TS-120S



TS-180S with DFC*



TS-180S, featuring DFC *(Digital Frequency Control, with four digitally tuned memories), is an all solid-state HF transceiver specially designed for the DXer, contest operator and other Amateurs who enjoy working the 160 through 10-meter bands. Also available is a full line of matching accessories, including the PS-30 base-station power supply, SP-180 external speaker with selectable audio filters, VFO-180 remote VFO, AT-180 antenna tuner/SWR and power meter/antenna switch, DF-180 digital frequency control, YK-88CW filter and YK-88SSB filter. Stand-up microphone shown above is optional.

TS-120S, popularly referred to as "a big little rig," is a compact, up to 200 watts PEP input, all solid-state HF transceiver with such standard features as built-in digital readout, IF shift, new PLL technology and requires no tuning. It's ideal for car or ham shack use. Accessories available include the PS-30 base-station power supply, SP-120 external speaker, VFO-120 remote VFO, AT-120 antenna tuner/SWR meter, MB-100 mobile mounting bracket and YK-88CW filter. Hand microphone shown above is optional.

See the exciting TS-180S and TS-120S at your Authorized Kenwood Dealer!

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