Solid-state displays are edging their way into new instrument designs. They have very definite advantages, such as increased reliability and extreme ruggedness. And their prices are becoming increasingly attractive. There are trade-offs involved, though, for both instrument designer and user. For details, see pgs. 24 and 51.
Stay thin with Dale

Honeywell Series 32 Computers Use Dale Thinline Connectors to Link Processing Options

Dale's EBTL 050 gives you the lowest profile available in a .050" edgeboard connector. This is one reason Honeywell uses it to interconnect processing options for its Series 32 Computers. Reliability is another reason. Connector failure in this application could shut down the entire system and risk data loss. According to Honeywell, the EBTL 050 "combines minimum insertion force with maximum withdrawal" and withstands low frequency vibration testing without discontinuity of more than 50 nanoseconds. And the cost is as low as the profile. Keep your package, and your budget, thin—call Dale for your next .050" requirement.

Phone 402—564-3131 for details or write for Connector Catalog.

DALE ELECTRONICS, INC.
Box 180, Yankton, South Dakota 57078

In Canada: Dale Electronics Canada, Ltd. A subsidiary of The Lionel Corporation

EBTL 050 THINLINE SERIES
Contacts: 8, 16, 20, 25, 32, 50 or 64 per side on .050" centers
Current Rating: 0.5 amp.
Board Thickness:
1/16" (.056"-.068")
1/32" (.027"-.035")
Profile:
.190" (1/16" boards)
.158" (1/32" boards)
Body: Glass-filled phenolic.
R.M.S. VOLTS -- the scale says -- but what about the circuits behind that scale?

All of us have been making rms readings of ac voltages for years. We know we have, it says so right on the front of the meter.

If someone were to ask what we mean by rms voltage, we could quickly explain the concept of "root mean square." In the interest of accuracy we might add that the rms voltage indication on most meters is true only for a sinusoidal wave. Unfortunately, most measurements are not made on true sinusoidal waves. However, for many applications, average responding meters are adequate.

But it would seem logical, where accuracy is important, to use a meter that measures true rms voltage no matter what the wave shape—a true rms voltmeter.

Why isn't this done more often? Well, until recently, most true rms voltmeters were expensive, limited in capability and rather slow responding.

Now Hewlett-Packard has adapted the thermocouple concept used in standard laboratories; added protective amplifiers to insure overload protection (800 V p-p); and reduced final-value step function response to less than 5 seconds.

When you combine these features with a low price of $600, it adds up to the HP 3400A—the first practical true rms voltmeter for general use in the 10 Hz to 10 MHz range. And, a high crest factor (ratio of peak to rms) allows you to measure noise and other non-sinusoidal wave forms at a ratio of 10:1 full scale or 100:1 at 10% of full scale. You get accurate noise and pulse measurements — without having to make non-standard corrections.

The 3400 isn't just a fine true rms voltmeter — although that's plenty in itself. It can also be used as an ac/dc converter and a current meter. Typical dc output accuracy is 0.75% of full scale from 50 Hz to 1 MHz. Use the HP 456A AC Current Probe ($250) and you get quick dependable current measurements. The 456A probe has a 1 mA to 1 mV conversion allowing direct readings up to 1 amp rms.

So, if all your measurements aren't made on true sinusoidal wave shapes and if you like direct accurate rms voltage indication no matter what you're measuring, it's time to check into the HP 3400A true rms voltmeter. For more information, contact your local HP field engineer. Or, write to Hewlett-Packard, Palo Alto, California 94304. Europe: 1217 Meyrin-Geneva, Switzerland.

INFORMATION RETRIEVAL NUMBER 2
Siemens, a world leader in the design and manufacture of sophisticated telecommunications and computer systems, has also become the world's leading producer of linear, memory and microwave ferrites. Ferrites that are performance and reliability engineered to meet our demanding system requirements.

Siemens pioneered T38 with permeabilities of 10,000, T9 and T10 high-density ferrites for recording heads, SM6 and M6 filter inductors for high packing densities, 12 mil extended temperature cores, plated wire memories and CVB7 microwave material.

Siemens design engineers are ready to assist you in solving your ferrite problems.

Siemens Corporation, 186 Wood Avenue South, Iselin, N.J., 08830.
(201) 494—1000.
Siemens. A three billion dollar name in quality products.
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Cover: Photo of Hewlett-Packard light-emitting-diode DPM display by Henry Ries

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Our new DOS brings batch processing costs down to $819 a month.

If you've been hanging on to old-fashioned ways because you thought a computer was too expensive, think again. Our new Disc Operating System brings the cost of computation and general purpose processing right down to where your budget lives.

With our new DOS, you'll easily create, check out and run your own programs. Use it for scientific calculations, business-accounting functions, information retrieval, inventory control, school administration — in fact, problem-solving of all kinds.

Anyone who can poke a typewriter key or pencil-mark a card can use our DOS. Because compilation, loading and execution are under the control of a teleprinter keyboard or batch input device.

On the other hand, if you're already batch processing with another system, give this a thought. Our DOS can probably do everything you're doing now — for about half the cost.

Because both the software and the hardware are fully modular, our DOS accommodates the needs of many different applications. Lets you vary the number of input/output devices. Add more core memory. Use a card reader as well as teleprinter. Add a line printer, paper tape punch, photo reader and magnetic tape. Other advantages include software protection and program segmentation. Plus automatic program retention so your programs can be easily reused.

Our basic DOS includes an 8K computer with direct memory access, 2.4 million-character disc, one teleprinter and one high-speed paper tape reader. Price is just $35,600. Or $819 per month on a five-year lease. And it's upward expandable for your future needs.

Get the full story by calling your local HP computer specialist. Or write to Hewlett-Packard, Palo Alto, California 94304; Europe: 1217 Meyrin-Geneva, Switzerland.
How to Buy a Good Power Supply
Without Spending a Bundle...

Take a long look at the Abbott line of over three thousand standard models with their prices listed. The unit shown above, for instance, is the Abbott Model B5S, a 60 Hz to DC converter which puts out 5 volts of regulated DC at 0.15 amps and sells for only $83. Other power outputs from 2 to 240 watts are available with any output voltage from 5 volts to 3,650 volts, all listed as standard models in our catalog. These power supplies feature close regulation, short circuit protection, and the latest state of art specifications for solid state modules.

If you really want to save money in buying your power supply, why spend many hours writing a complicated specification? And why order a special custom-built unit which will cost a bundle—and may bring a bundle of headaches. As soon as your power requirements are firmed up, check the Abbott Catalog or EEM (see below) and you may be pleasantly surprised to find that Abbott already has standard power supplies to meet your requirements—and the prices are listed. Merely phone, wire, or write to Abbott for an immediate delivery quotation. Many units are carried in stock.

Abbott manufactures a wide variety of different types of power supply modules including:

- 600 VDC to DC, Regulated
- 400 VDC to DC, Regulated
- 28 VDC to DC, Regulated
- 28 VDC to 400 VDC, 1φ or 3φ
- 24 VDC to 600 VDC, 1φ

Please see pages 930 to 949 of your 1970-71 EEM (ELECTRONIC ENGINEERS MASTER Catalog) for complete information on Abbott modules.

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INFORMATION RETREIVAL NUMBER 5
letters

Engineers don’t starve—they only lose prestige

Sir:
In observing the present crisis in engineering employment, especially in the aerospace and defense industries, two alarming facts have become apparent. Going hungry is not one of these ugly facts. I maintain that an engineer has the ability to do damn near anything. He may sweat in times like these, but he won’t starve!

Far more important than our own financial security are the future of America’s technical foundation and the prestige of the engineering profession.

The glaring facts are these:
1. The United States’ defense establishment, founded on technology, is being destroyed from within. Technology must be continually moving forward, just to keep our nation’s strength equal to the task of maintaining our freedom and preventing aggression against the free world. In these matters, to stand still is to die, and to willfully slide back, as we are doing today, is suicide. As engineers are forced out of work and find more secure, easier, and sometimes more enjoyable ways of making a living, they are forever lost to our future technical manpower needs.

2. The prestige of being a professional engineer is fast becoming a thing of the past. Engineers are individuals, not members of a tough-minded organization of responsible people concerned with the protection of the profession and its members. Therefore, companies are taking advantage of this to fire people, deny cost-of-living and merit raises, cut salaries, demand free overtime from engineers, and subject them to degrading systems of surveillance to detect anyone who isn’t eagerly doing more than his share.

Many careers are being destroyed and engineers are being taken advantage of today by management. We of the profession, if we are to deserve the prestige of being called professional individuals, must exercise our responsibility to our nation, its people, and its future. We must make certain that the present irresponsible dismantling of the foundations of our national existence and the undermining of our technical expertise and professional prestige are stopped right now!

A Concerned Engineer

Editor’s note: In asking us to withhold his name, the writer says, “I believe a letter signed by one unknown engineer will not be as effective as one signed as above, who could be anyone and who may well be speaking for everyone.”

Bureau of Standards doesn’t test products

Sir:
“Clearing” a dozen or more makes of microwave ovens, reportedly tested by the National Bureau of Standards (see “Capital Capsules,” ED 25, Dec. 6, 1970, p. 46), should have been attributed to the Bureau of Radiological Health of the Department of Health, Education and Welfare. Although the NBS is developing monitoring instruments and measurement techniques for BRH, it does not pass judgment on the acceptability of specific products.

Jim Kluge
National Bureau of Standards
Boulder, Colo.
four interface problems:

1. Eliminate interface maintenance between a telegraph line and TTL logic for $3.55.

2. Trigger a remote SCR from TTL logic for $4.50.

3. Eliminate ground loop spikes between computer and remote terminal for $3.95.

4. Couple TTL logic to 1.5 kV CRT blanking grid for $3.95.

four opto-isolator solutions:

1. GaAs infrared LED
   - $t_r = 40 \text{ ns} \ @ \ R_L = 50 \Omega$
   - $I_F < 100 \text{ mA cont.}$
   - $V_F = 1.3 \text{ V}$
   - $V_{BR} = 3 \text{ V}$
   - $\lambda = 9000 \text{ Å}$
   - This emitter makes solutions 2, 3, and 4 possible.

2. Silicon photodiode MCD 2
   - $t_r = 110 \text{ ns} \ @ \ R_L = 100 \Omega$
   - $f_{fM} = 200 \text{ kHz} \ @ \ R_L = 100 \Omega$
   - $V_{BR} > 75 \text{ V}$
   - $I_L/I_F = 0.2\% \text{ typ.}$
   - Solves 3 and 4 above.

3. Silicon phototransistor MCT 2
   - $t_r = 2 \mu s \ @ \ R_L = 100 \Omega$
   - $f_{fM} = 200 \text{ kHz} \ @ \ R_L = 100 \Omega$
   - $V_{BRCEO} > 30 \text{ V}$
   - $I_{C}/I_F = 35\% \text{ typ.}$
   - $I_{C\text{sat}}/I_F = 5\% \text{ typ.}$
   - Solves 1 above.

4. Silicon photoSCR MCS 2
   - $t_{fM} = 5 \mu s \ @ \ I_F = 20 \text{ mA}$
   - $I_A < .15 \text{ A cont.}$
   - $V_A = 200 \text{ V max.}$
   - $I_F \text{turnoff} < 10 \text{ mA}$
   - Solves 2 above.

...and an answer kit: $9.95.

The opto-isolator solves tough design problems. To let you work with these new devices, we’ve put together an Opto-isolator Answer Kit. Contains an MCT 2, MCD 2, MCS 2 in our six-lead Iso-DIP package at about half the price of the discrete parts — plus a new volume of GaAsLITE Tips that shows how to design opto-isolators into your problem circuits. Order from your Monsanto distributor or write Monsanto Electronic Special Products, 10131 Bubb Road, Cupertino, CA 95014. (408) 257-2140.

All part prices are suggested resale price in 1,000 quantities.

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For complete technical information on all 10 of our fast response, wideband op amps, contact your Burr-Brown Engineering Representative or use this publication's reader service card.

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<td>3400A</td>
<td>Industry's fastest differential op amp. 100 MHz bandwidth. 1000 V/μsec slew rate. 1 μsec settling to .01%. FET input. Price (1-9) $55.00</td>
<td>Video and pulse amplification.</td>
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<td>The industry's only fast settling op amp with guaranteed accuracy as a noninverting buffer. Gain accuracy better than .01%. 1 μsec settling to .01%. FET input. Price (1-9) $42.00</td>
<td>A/D, D/A converters... and multiplexers.</td>
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<td>Fast settling, 1 μsec to .01%. Differential FET input. Low Cost, (1-9) $37.00</td>
<td>Ideal for high-speed precision D/A converters with monolithic quad switches.</td>
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<td>Slew rate of 1000 V/μsec. Bandwidth, 50 MHz. Output current, ±100 mA. Price (1-9) $69.00</td>
<td>High speed line driving</td>
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RCA
Solid State Division

MARCH 1971

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March 22-25

CIRCLE NO. 405

March 31-Apr. 2

CIRCLE NO. 406

April 12-15

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And if you’d like complete information on all three Friden digital printers, ask Mr. Gary Dotzler, Sales Manager, OEM Products, Friden Division, The Singer Company, San Leandro, California 94577.

*Singer Friden Division

INFORMATION RETRIEVAL NUMBER 13

INFORMATION RETRIEVAL NUMBER 14
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Computer Microtechnology Inc. P. O. Box 7050, Sunnyvale, Calif. 94086

INFORMATION RETRIEVAL NUMBER 15
RF.
If You Get The Message
From The Data Sheet,
It’s A Signal You’ll Get Through.

Ever wondered what “infinite VSWR” is? Or why many RF transistors don’t conform to JEDEC-registered specs like they should? Or what a “typical” power-out spec is?

Lots of data sheets keep you wondering. Ours don’t.

We spec everything from input/output resistance and capacitance (y & s parameters) for small signal stuff to impedance parameters for large signal devices. Like good blueprints, our data sheets tell you everything you need to know about the RF device you’ve selected so traditional trial-and-error is minimized. And so is your design’s failure.

We call them True Blueprints around the factory.

You can take a realistic view of Motorola RF devices because we do. Load mismatch, depending on the unit, is usually performed at a 10:1 or 30:1 VSWR. One that’s testable. Acceptable. And designable.

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You’re assured each device meets a registered minimum performance standard and is therefore subject to production controls. The one you buy next month will perform like the one you bought last month.

You’re even given fine points like leakage current, sat and emitter-base breakdown voltages just because JEDEC says that’s what you should be given.

We want to give you something else, too: 10% off your next order for up to 999 of any Motorola RF small signal or RF power transistor to prove our devices are as good as our data sheets. Fill in and mail the coupon to your local Motorola distributor. He’ll get through to you immediately to take your order.

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Some of the case styles in which Sprague DST Pulse Transformers are available. Note the in-line leads.

You can select the transformer design you need from the new Sprague DST Family, a fully-characterized series of Designer Specified Transformers which Sprague Electric has pioneered. It's easy. Start with the two basic parameters dictated by your circuit requirements: primary (magnetizing) inductance and volt-second capacity.

New Sprague engineering data gives basic information from which all nominal sine wave parameters are derived. This data allows you to specify the one transformer from thousands of possibilities which will optimize performance in your application.

Design Style A minimizes magnetizing inductance change as a function of temperature. Typically it's < ±10% change from 0 to 60°C; < ±30% from -55 to +85°C.

Design Style B and C give you broad band-pass characteristics, and still keep magnetizing inductance change < ±15% from 0 to 60°C.

Design Style D is fast. Associated leakage inductance and coupling capacitance are kept at a minimum. This style is just what you need for interstage and coupling devices in computer drive circuits.

The Sprague DST Series packs a lot of transformer into minimum volume packages — epoxy dipped for minimum cost, or pre-molded. The 100 mil in-line lead spacing is compatible with integrated circuit mounting dimensions on printed wiring boards.

A new dual-two-stage filter that is able to use YIG-tuning techniques and still be compatible with microwave integrated circuits is now on the market. The unit has two separate two-stage filters within the same magnetic structure, yet measures a mere 0.75 inch in diameter and 2.1 inches in length.

Designed for use in the S band from 2 to 4 GHz, the new YIG-tuned filter is adaptable to two, three, and four-stage filter configurations in addition to the two-stage designs mentioned. It can also be used as the magnetic structure for YIG-tuned oscillators.

You can divide clock frequencies by any integral divisor by selectively controlling the input signals to a common pulse-gated ripple counter. And only one monostable-gate generator is needed—an advantage that is most apparent for large input/output frequency ratios that would otherwise require many external gates.

Either TTL or RTL logic can be used for the circuit, and the method is easily adaptable to MSI or conventional manufacturing. For example, 10 flip-flops and four dual gates, either deposited on one MSI chip or assembled from ICs onto a circuit card, can provide a low-cost, adjustable divider to cover the entire audio range from one clock frequency.

Of the many challenges to the Nixie tube for the title of king of digital readout devices, the strongest is the light-emitting diode (LED).

Significant space savings for designers, a lifetime of 10 years or more and TTL compatibility are the main features being offered.

And prices, while generally higher than those for Nixies now, are reported to be headed downward. At least 10 LED manufacturers have already entered the fray. Some offer dot-matrix displays and others seven-segment displays, in digit sizes from 1/8 inch to 0.85 inch. The package may be dual in-line, flatpack or edge-mounted.
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DATA DISPLAY PRODUCTS

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Over-the-counter market gets an electronic ‘board’

The over-the-counter stock market now has its own “board”—an electronic one. A nationwide computer system for reporting quotations, averages and volumes on 2300 over-the-counter stocks went into operation this month.

Built and operated by the Bunker-Ramo Corp. for the National Association of Securities Dealers, the electronic “stock exchange” consists of 40,000 CRT terminals and keyboards at more than 700 over-the-counter market-maker and stock-retailer sites. In its computer center at Trumbull, Conn., Bunker Ramo has two Univac 1108 computers, either of which is capable of operating the entire system by itself. Three banks of core totaling 196-k words are used to store the stock quotations.

From each of the terminals, market makers enter and update their quotes and instantaneously interrogate the central computers to obtain the quotations of all other market makers on any given issue in the system. Previously all of this communication was done by telephone.

The electronic system has a capacity of 20,000 stocks. Unlike the New York and American Exchanges, the new over-the-counter system provides more than one bid and ask quotation an each issue. If 10 market makers are quoting on an issue, the quotes of all 10 are shown on the subscribers terminal.

Bunker Ramo estimates that the system will handle 56 entries and interrogations per second—more than one million calls a day.

The computer center at Trumbull Conn., is linked by high-speed, dedicated data lines to subcenters in New York, Chicago, San Francisco, Dallas and Atlanta.

Microwave radio system links local police to FBI

A microwave radio communications system that uses a minicomputer to store and switch messages will be installed in San Mateo County, Calif., next week.

It will link all police departments in the county to each other and, by way of Sacramento, to state data bases with information on stolen cars and gun registrations. The Federal Bureau of Investigation records in Washington, D. C., will also be accessible, providing data on 2.5-million active records including wanted persons lists.

The key to the system is a new Motorola MDP 1000 minicomputer. This machine stores messages received over the microwave link from local police offices and forwards them to addressed agencies at 100 words per minute.

William D. Goheen, communications supervisor of San Mateo County, explains that in other police communication systems a clerk at the initiating agency must cut a tape, dial the address and then send the message at 100 wpm by tape. This requires a trained typewriter operator.

Ultimately Goheen hopes to have a terminal with a standard typewriter keyboard and CRT display in every police car. Patrolmen will then be able to access the switching computer by radio link and thus have fingertip access to data bases throughout the state and in Washington, D. C. Such a terminal is already available, Goheen says, from Sylvania Electronic Systems, Mountain View, Calif.

Navy’s new subs to get improved electronics

Electronic subsystems for the Navy’s SSN 688 nuclear, high-speed, attack submarines will be a big step ahead of subsystems on the Sturgeon-class attack sub, now in use.

The new underwater fleet will be equipped with an improved sensor system (IBM has a $50-million contract for five of these sonars); an improved acoustic countermeasure system; advanced fire control; a central computer complex (as opposed to the minicomputer-for-every-subsystem concept); improved navigation capability, and an advanced electronic countermeasures receiver.

The new submarines will carry dual-purpose versions of the wire-guided Mark 48 torpedo—dual in that it is capable of destroying both submarines and surface ships. Two dual-purpose Mark 48s are being built. Clevite is building one, the Mod 1, and Westinghouse the other, the Mod 2.

The first 12 ships will cost an estimated $1.9-billion. The Newport News (Va.) Shipbuilding and Dry Dock Co. will build the lead ship and four other submarines, and General Dynamics Corp.’s Electric Boat Div. in Groton, Conn., will build seven.

Digital control center cuts software costs

A major factor in the high cost of digital process control equipment is the cost of developing and maintaining programs. A present rule of thumb is that a manufacturer designing a new system should ex-
pect to spend a dollar for software for each hardware dollar he spends.

Fisher Controls Co., of Marshall-town, Iowa, has come up with a digital control center—which it calls dc3—that reduces the software cost to the user to 1/10th that of the hardware. This has been accomplished, according to J. B. Duncan, dc3 product manager, by the use of a read-only memory in conjunction with a new real-time process control language called dc3—which is derived from the BASIC language.

This means that the compiler is in the hardware and is always available for use by the programmer. The executive or monitor programs are written by the user as he develops his programs.

For a control project involving 25 direct digital control loops and a normal number of other functions, such as discrete outputs and alarming, the dc3 price would be about $50,000. The software costs for this project would be typically $5000, Duncan says. Deliveries begin in March.

Electron-beam mining to be tested in field

Is there a future for electronics in mining operations? The U. S. Interior Dept.'s Bureau of Mines has decided to find out. It has awarded a contract to Westinghouse Electric Corp., Pittsburgh, to field-test an electron-beam rock excavation machine that the company has developed.

The 36 kW machine makes use of a high-energy electron beam to melt deep but narrow cuts in rocks. The beam is produced without the aid of the huge vacuum chamber heretofore used with most high-energy electron beams. Instead the beam is passed through a series of small chambers, each separated by an aperture of 1/10th of 1/20th of an inch in diameter. Each chamber is pumped to preserve a partial vacuum. In the process, the beam is focussed electromagnetically and delivered to the surface where it is needed.

The electron-beam method has the potential of cutting the cost and increasing the speed of drilling and tunneling through rock.

The tests will be supervised by Westinghouse's Missile Launching and Handling Dept. in Sunnyvale, Calif., supported by the company's research laboratories near Pittsburgh. Test operations will be carried out by the Hi-Z Mining Corp. of Albuquerque, N. M. The $240,000 contract is being funded by the Defense Dept.'s Advanced Research Projects Agency.

1500-W power plant to be unfurled in space

The first satellite with a roll-up solar-cell array—a 1500-W power plant—is slated to be launched by the Air Force in October. The array will be rolled up on a 10-inch drum during launching.

The satellite was developed for the Air Force Systems Command, Wright-Patterson AFB, Ohio, by Hughes Aircraft Co., El Segundo, Calif.

According to David Massie, project engineer for the Air Force Aero Propulsion Laboratory, the new design will be 50% lighter than traditional rigid, folded-wing solar arrays, and it will be smaller in volume by 50 to 75%.

When the satellite is in its 400-mile-high polar orbit, the solar cells will be extended as two 16-by-15-1/2-foot panels. Massie says the satellite is an experimental one that will be used to test the concept of developing power plants of 5000 to 10,000 W for future military satellites. Most satellites have solar cells that produce 100 to 1000 W.

The satellite has four electron subsystems—one for extending the solar array, another for acquiring and tracking the sun, a third for controlling the power and energy storage, and the fourth for operating the instrumentation.

Underwater holograms tested in Navy sub

An acoustic holographic system has been installed on a Navy deep submersible vehicle and for the first time has displayed underwater objects in three dimensions in real time.

The system, which has a range of several hundred feet, was developed by Bendix Research Laboratories, Southfield, Mich., under a contract with the Office of Naval Research.

There are six major components in the Bendix system:

- a sound transmitter
- a receiving array of hydrophones.
- signal processing equipment to convert acoustic signals to optical signals.
- a laser to reconstruct the optical signals to a true image in real time.
- a closed-circuit TV monitor on which the image is displayed.

The Navy says the system can be used for rescue operations and bot- tom topography mapping.

'Silicon Valley' gets new semiconductor firm

"Silicon Valley's" first new semiconductor company for 1971 is Antex Industries, Inc., in Palo Alto, Calif. It was founded by Ken T. Chow, former president of Electro-Nuclear Laboratories, Menlo Park, Calif., a manufacturer of silicon photocells and detectors.

The new company will design and manufacture custom MOS devices for watches and calculators. It will also make gallium arsenide single crystals for the light-emitting-diode industry.

Laser to spot fog banks and alert Coast Guard

A laser device is being developed by the U. S. Dept. of Transporta-
tion for detecting fog banks a half mile to three miles away.

The Coast Guard, which is to use the device, has Videograph equip-
ment at present. This detects general fog conditions but not individual fog banks.

The new system would operate by beaming the laser light outward from a shore installation. As the light beam reached a fog bank, it would trigger alarms, such as fog horns and buoy signals, in unmanned lighthouses.

Studies are continuing on ranging the fog-detection device so that it not only detects the fog bank but determines its distance from shore. The work is being carried out at the Dept. of Transportation Center in Cambridge, Mass.
HERE ARE TWO EASY WAYS TO SOLVE LIGHTED PUSH BUTTON SWITCH PROBLEMS. Economically. Reliably. Fast. The Molex 1175 snap mounts. Offers spade or wire terminals for fast, easy assembly. A choice of nine colors, 500 variations. And look at the Molex 1820. You can use one, or a gang of them, for an infinite variety of applications. Lighted push button can be wired to light independently of the switch. And it's available in colors galore. Best of all . . . both switches are priced considerably under one dollar in quantity. These components are good examples of the Molex creative approach to design problems. And we have the ability to design reliability and ease of assembly into a product without letting costs run wild due to over-engineering. If this makes sense, and you would like a free sample of either the 1175 or 1820 switch, write: Molex Incorporated, Downers Grove, Illinois 60515. Or phone (312) 969-4550.

...creating components that simplify circuitry
LED: The No. 1 challenger to Nixies in digital readouts

Of the many challenges to the Nixie tube for the title of king of digital readout devices, the strongest is being mounted by the light-emitting diode (LED).

Significant space savings for designers, a lifetime of 10 years or more and TTL compatibility are the main features being offered.

And prices, while generally higher than those for Nixies now, are reported headed downward. A figure of $10 to $11 per digit is fairly common at present. But in very large quantities—10,000 to 100,000 digits—prices on the order of $3 for a 1/4-inch-high digit have been quoted. In 18 to 24 months, say several manufacturers, including Aaron Kestenbaum, president of Opcoa in Edison, N. J., the price of LED readouts in 1/4-inch digits should be down to $1 to $2 per digit for large-quantity orders.

At least 10 LED manufacturers have already entered the fray. Some offer dot-matrix displays, others seven-segment displays. Some offer 1/8-inch digits, others 1/4-inch and still others digits as large as 0.85 inch. The package may be dual-in-line, flatpack or edge-mounted.

Present Nixie prices, according to a spokesman for Burroughs of Plainfield, N. J.—the producer of the trademarked tubes—range from $6.75 to $45 per digit. In large quantities—50,000 to 100,000—the Nixie price is typically anywhere from less than $3 to $7, the spokesman reports.

Nixies range in digit height from 3/10ths inch to 2-1/2 inches. "Almost all" the Nixies sold today, Burroughs says, are the "long-life" type, good on the average for 100,000 hours or 12 years.

Other gas-discharge tubes that compete with Nixies offer comparable price, size and performance.

The basic material from which the LED is fabricated is still the subject of disagreement. Some manufacturers say that gallium arsenide phosphide (GaAsP) is best; others like gallium phosphide (GaP).

GaAsP proponents note that its wavelength is closer to the optimum sensitivity of the human eye—it has better "luminous efficiency" than GaP. Moreover more research has been done on the production of GaAsP than on GaP; manufacturers know more about the material.

In favor of GaP is this strong argument: Potentially it's cheaper to manufacture, because liquid-phase epitaxy can be used instead of vapor-phase epitaxy. The GaP epitaxial layer can be grown in about five minutes, as opposed to one day for the epitaxial layer of GaAsP. The GaP readout also holds promise of a green readout, which is more sensitive to the eye than the red that present LEDs admit.

Only one manufacturer is using GaP at present, however: Opcoa. The process was developed originally at Bell Telephone Laboratories, and Opcoa's present engineering team was formerly with Bell.

What does the designer have to know about LEDs to select or design instruments with digital displays? First, he must know the tradeoffs.

On the plus side:

- LEDs have rapid rise times (in nanoseconds), making it very easy to multiplex several digits off a single decoder driver. This permits the display of several digits with the use of only enough power to drive a single digit continuously.
- LED displays are single-plane devices. Whereas the Nixie is constructed with 10 cathodes, one behind the other, LED digits are all generated on a single flat plane. The result is a significantly greater viewing angle for the LED.
- LED displays are small, flat and thin. They can be designed into portable, compact instruments, or space can be left for additional circuitry in bench instruments. Nixies generally take up much more space than LED devices.

On the minus side:

- The character size of LEDs is generally quite small. If a readout must be clear from a distance of 25 feet or more, Nixies or some other form of large readout must be used. Tests with 1/4-inch LEDs have shown that the maximum reading distance is 10 to 12 feet.
- When LEDs fail, the display may black out suddenly and completely. Nixies usually fail gradually—the elements put out less and less light once they start to degrade.
- Nixies provide a curved digit that is very readable. The seven-segment digits in LEDs are somewhat boxlike in character, and when the digits are changing, they look like 8s. The digits in dot-matrix LEDs are better formed, but still not as complete as the Nixie digits. In addition either circularly polarized, narrowband red or smoked black filters are recommended over the displays of all LEDs to mask reflected light and to make them more readable.

(continued on next page)
Type 707 Spectrum Analyzer, made by AIL Div. of Cutler-Hammer, Deer Park, N. Y.

Model RDS-16 Desk Pollution Monitoring System, made by Time-Zero Corp., Hawthorne, Calif.

Type 3330 Digital Panel Meter, made by Digilin, Glendale, Calif.

Instruments with LED displays

Model 3431A Digital Panel Meter, made by Hewlett-Packard Co., Loveland, Div., Loveland, Colo.
Because of their rapid rise times, LEDs have no persistence. When the current is turned off, the light output ceases in nanoseconds.

Multiplexing with LEDs

A major design consideration with LEDs is that of multiplexing several off a single decoder driver.

Strobe rates of 1 kHz and higher are recommended. Since the human eye tends to be a peak detector, low-duty-cycle, high-peak current pulses can be fed into LEDs giving a high-intensity display.

If a single decoder driver is used with each digit, the current should be kept under 50 mA, or the junction will heat up and the LED will go into a saturated state. If this occurs, an increase in current will not yield an increase in light output.

At lower current levels, or in short pulses, light output is proportional to the input-drive current. In any case, current-limiting resistors are necessary on the cathode pins for burnout protection.

Several instruments with LED readouts—spectrum analyzers, calculators, various types of monitoring consoles, digital panel meters and digital counters—have already reached the market, and many more are expected. All of Hewlett-Packard's new digital voltmeters, for example, will have LED displays from now on. The first H-P instrument with a LED display, a digital panel meter, has just hit the market (see "DPM Uses LED Display and Custom Parts for Reliability," p. 51 this issue of ED.)

Hardware for the 'cashless society'

And Video Systems Corp., Pennsauken, N. J., displayed its credit-card validation and inventory control systems last month at the National Retail Merchants Association Convention in New York City.

At the same convention, Concord Computing of Bedford, Mass., demonstrated its Validier, which operates through a Touch-Tone telephone. The system authorizes credit and processes sales slips.

Data banks containing credit information for millions of people are being formed, and the hardware and systems needed for accessing that information are being designed and marketed. This leaves the safeguarding of credit information largely to the discretion of sales clerks and cashiers.

The problems of making absolutely certain that the holder of a credit card is, in fact, the authorized owner is greatly facilitated by telephone-line transmission to a central data bank. But two Bell Laboratory engineers, Dan Miller, and Terry Prince, agree that secret user codes and even voiceprint comparisons by telephone may eventually find a place in the "cashless society."

The Bell System's plastic credit card meets size requirements set by the American Standards Institute, New York City. The card can be used not only for ordering merchandise but also "to verify bank balances or pay bills," Miller says.

The credit-verification systems use a Touch-Tone telephone or a specially designed keyboard to input information to a central computer. The computer replies either by synthesized speech or a special tone indicating that credit is approved or denied. The amount of each transaction, the date and personal identification are recorded and numbered.

The TRW credit authorization system will be used by a number of banks, supermarkets and national department store chains, including Arnold Constable, Montgomery Ward, the May Co., Broadway, Orbach's and Nordstrom Best.

The Video Systems Corp. equipment is a desk-top video terminal with alpha-numeric keyboard and seven control keys for entering, deleting and updating information. More in the nature of a mini-computer this system—called the Inventron—is for controlling inventory, showing work in progress, and displaying back-order lists to sales managers as well as providing customer credit and credit card validation.

The Validier uses a wide-area dedicated communications network and can be installed in increments, starting with a single department and expanding to an entire store or chain. A typical credit transaction takes less than 15 seconds to complete, according to Concord Computing Co.
New job for the busy computer:
Designing artificial heart valves

Designing an artificial valve for a heart patient can now be done by computer. The cardiovascular system varies from person to person, and the designer's conventional approach is a painstaking cut-and-try method.

But the new computer techniques developed at the University of Utah's Computer Sciences Div. hold promise of easing and speeding the work. The techniques are useful not only for designing artificial hearts and valves; they can be used to study the flow of blood through arteries. It's hoped that the latter use will prove beneficial in research on atherosclerosis.

The experimental approach

Without the computer, the designer of an artificial heart valve must build a Plexiglas model of the patient's aorta with the valve in it. Then a fluid that contains visible particles is run through the model, and the designer looks for areas of turbulence—that is, areas where the velocity of the fluid flow is changing constantly in both magnitude and direction. This is a danger sign, telling the designer to modify the shape of the valve. Excessive turbulence in a real aorta produces lesions in the artery wall, leading eventually to strokes or internal hemorrhaging.

With the computer, the shapes of the aorta and the valve are outlined on a CRT screen, and turbulence is indicated by large dense spots. The motion of the valve back and forth in the aorta is observed on the screen. Furthermore the shape of the valve can be changed instantaneously by a light pen and the change in turbulence studied.

Dr. Harvey S. Greenfield, research professor at the University of Utah, Salt Lake City, explains that the technique consists basically of entering the general equations of fluid motion into the computer, having it solve them for particular valve and artery shapes and sizes, then displaying these solutions on the CRT.

The figures below for example, are drawings of views on the CRT screen. The diagram at left shows a longitudinal cross-section of the aorta, with a disk-shaped valve in the left end. The valve moves on the screen from the full open position shown leftward to a closed position in the narrow passageway. The lines along the walls of the aorta indicate the direction of motion of the blood, and the black spots indicate turbulence. The drawing at right shows a three-dimensional plot of the turbulence. The height of the peaks gives an approximate indication of the amount of turbulence.

New research tool possible

Dr. Greenfield says that CRT displays of a valve moving back and forth in the aorta show excessive turbulence in the same areas as Plexiglas models that contain similar valves.

Furthermore, he notes, the computer displays of blood flow through the arteries indicate excessive turbulence in the same areas where surgeons consistently find lesions in atherosclerosis patients. He anticipates, therefore, that computer-aided design will be a useful technique for studying atherosclerosis as well as for designing artificial heart valves and hearts.

Researchers at the University of Utah plan eventually to use a head-mounted display system (see "Computer Constructs 3-D Castles in the Air," ED 19, September 13, 1970, p. 32) that will allow the viewer not just to observe a picture of a moving heart valve or diseased artery on the CRT, but actually to "step inside" the artery and view the theoretical "blood" as it flows toward him.
Introducing the high-low POWER SUPPLY (high performance—low profile)

A 16-stage deflector capable of steering a laser beam into a raster of 65,536 half-overlapping positions has been introduced by Philips Research Laboratories of Hamburg, Germany. It permits random addressing at a rate of 250 kHz, and the beam may be switched from position to position in 0.2 ns. Philips spokesmen claim that a 5-by-5-foot display produced by the new system can be seen in daylight. Light losses in the system are only about 10%.

A laser system that can measure flow rates from 1 kilometer per second to 0.01 millimeter per second has been devised by scientists at Brown Boveri Research Center in Baden, Switzerland. The technique can be used with any transparent liquid, provided it contains diffractioning particles. The system involves focusing a He-Ne laser to a fine point through a glass tube and into the liquid whose velocity is to be measured. Readings are taken of the beam frequency from three perpendicular directions to compute the total velocity vector.

A traveling-wave cathode-ray tube with an extremely high (0-to-6 GHz) bandwidth has been perfected by the Philips Research Laboratories at Val-de-Marne, France. The device can reproduce extremely brief nonrecurring events. The transient signal is impressed onto a helical delay line that also serves to deflect the tube beam current. The signal travels at substantially the same speed as the electrons composing the beam current. This means that individual electrons experience a constant magnetic field, and the signal is faithfully reproduced.

All the arithmetic and logic for Sumlock Comptometer's newly announced desk calculator have been squeezed onto just five microcircuit chips. Sumlock, a London-based calculator company, designed the metal-oxide microcircuits in collaboration with a team of engineers based at the Glenrothes (Scotland) facility of General Instruments Microelectronics. Although a few three-chip calculators have been produced by American MOS manufacturers—mostly for Japanese calculator companies—the Sumlock design approach remains a novel and low-cost solution to design problems. By multiplexing timing and control signals and keyed information onto a common bus, the pin count has been drastically reduced. Sumlock engineers were able to use 11-lead TO-5 cans costing only a few cents.

Low-cost ship-to-shore communications for merchant shipping via satellite has been proved feasible as a result of tests recently completed by the British Post Office in collaboration with the radio operating companies and suppliers of equipment. NASA and the U. S. Coast Guard also gave assistance. To cut costs, the normal dish antenna stabilized against ship roll was eliminated. In its place a crossed Yagi array was installed on the experimental container ship, Atlantic Causeway. The resulting 30-degree beam width kept the satellite in sight even during the most severe rolling.

The multi-role combat aircraft now being jointly developed by Great Britain, Germany and Italy is to be fitted with a low-light-level TV laser system. This will be jointly developed by Barr & Stroud, Ltd., and the British Aircraft Corp., two British firms, and Eltro Gmbh of Heidelberg, Germany. The low-light-level system will be used in reconnaissance, and the laser system will probably be an essential feature of an automatic ranging facility.
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See how much easier you could measure high-frequency waveforms on HP's new scope?

Get the big picture! The 180-Series oscilloscope family now has a "big brother"—the 182A—a scope with a viewing area 66% larger than any other high-frequency scope, and three times larger than some. Yet, despite its larger viewing area, the 182A takes up no more bench space than the 180A, and maintains full bandwidth of 180 System plug-ins to 100 MHz.

The result is easy viewing for you—even at a distance. The 182A's 8-div x 10-div CRT is marked off in big 1.3-cm x 1.3-cm squares. And internal graticules allow accurate readings from any angle—a real plus in systems-testing work.

This new big-screen CRT is possible because of HP's pioneering advances in CRT technology. Improved HP expansion-mesh magnification techniques used in the 182A CRT make it possible to have a big-screen scope with 100 MHz capabilities, while still retaining the sensitivity required for compatibility with solid-state vertical amplifiers. Thus, you get easy-to-interpret displays, even in 4-channel or TDR work.

Because the 182A is part of the 180 "family," it will take ten different 180 System plug-ins—up to and including the 100 MHz 1802A dual-channel vertical amplifier—without degradation. This means you can upgrade your existing system without having to replace any of your present HP plug-ins. It also assures compatibility in the future.

Price of the 182A mainframe is only $1100; plug-ins start at $525. For further information on the new 182A, or any element of the HP 180-System "family," contact your local HP field engineer, or write to Hewlett-Packard, Palo Alto, California 94304. In Europe: 1217 Meyrin-Geneva, Switzerland.
Tax break for business stirs mixed reaction

It's still too early to tell what effect President Nixon's relaxation of tax-depreciation regulations will have on capital spending in the electronics industry. Some manufacturers are wary that it will have any effect this year. Although tax benefits will be felt in fiscal 1971, these observers note that most companies have already planned their major purchases for this year. On the other hand companies that have deferred capital improvements may be encouraged to start them soon, other observers feel.

The Electronics Industry Association says its members are still studying the new regulations and are unsure how they might influence major purchases or the job market.

The changes in the tax law will permit businesses to cut the average depreciation "life" of machinery by 20% and to take a full year's depreciation on anything bought in the first half of the year and a half year's depreciation on purchases in the last six months. The "reserve-ratio" test, under which businesses formerly were required to justify the actual period of time during which equipment was used and its depreciation "life," also has been eliminated.

$15-million sought for new military network

The Defense Dept. is expected to ask Congress for about $15-million this year to fund Tritac, an Army-Navy-Air Force tactical communications network proposed as a replacement for Project Mallard, which died in the Senate last year. Mallard included foreign governments in the net as well as the three American military services, and this led many legislators to believe the project was so complex that it would be unworkable. There are still problems remaining, such as what type of computer operation is to be used in message switching.

Airlines accept Administration's uhf satellite plans

The Administration's decision to put up two uhf satellites for aviation use won't be disputed by the nation's airlines, despite the fact that they and the Federal Aviation Administration have preferred a vhf system in the past. A spokesman for the Air Transport Association told ELECTRONIC DESIGN that the carriers could live with the plan, as outlined by the Office of Telecommunications Policy. The satellites are to be operational by 1980. "At least we now have a policy," the spokesman said—something that has been lacking for almost a decade.

The Office of Telecommunications Policy is looking for the first satellite to be launched over the Pacific in 1973, with tests following, and a similar satellite over the Atlantic in 1975. The system is to operate near 1600 MHz and will involve automatic data collection, data processing, surveillance and traffic control, with displays utilizing digital data links and digital processing techniques. The Dept. of Transportation will run the system, with NASA and the State Department participating.
The airlines favored a vhf system because it matched existing equipment that they have and because it could be put into operation faster. But now, beset by financial woes, the carriers seem content to accept the uhf system and wait until some future date, when their money problems hopefully will be a thing of the past. NASA and European governments were the leading exponents of the uhf system.

$2-billion Increase in defense budget expected

Sources in the House Armed Services Committee indicate that the Defense Dept. budget for fiscal 1971-72 will come in at about $76-billion, compared with the $74-billion estimated for 1970-71. A part of the increase, the sources say, will be for increased payroll costs. But R&D spending is expected to rise sharply—possibly as much as a billion—from the $7-billion in 1970-71. Defense purchases are expected to remain about the same. A large part of the R&D increase will be for the B-1 bomber and new submarine programs.

NASA denies plans for robot moon exploration

Reports that robot moon-exploration plans are being worked on within NASA are denied by officials of the agency. The plans, according to reports circulating in Washington, will be formalized by this spring.

The robot programs supposedly resemble those that the Russians now have in operation, and the feeling is that if NASA did go to mechanical exploration of the moon it could, hopefully, lead to greater cooperation with the Russians. It might also, say the reports, take the budgetary heat off NASA from Congressional critics of the manned-flight program. But the furthest that manned-flight officials will go is to say that for a while they “toyed” with the idea of making the lunar exploration vehicle in dual mode—that is it could stay on the moon, move about, continue to scoop up samples and then later meet with another Apollo crew at a prearranged rendezvous. But this, they hasten to add, was dropped some time ago.

Capital Capsules: The Air Force now estimates that each of the 325 F-5-21 aircraft it plans to buy from Northrop will carry about $100,000 in off-the-shelf electronics. Flyaway costs of the aircraft are estimated at $1.6-million apiece... The Housing and Urban Development Department is working on a project to put 1500 unemployed aerospace technicians to work in the Model Cities program. Initial cost of the program is estimated at $4-million... The Federal Aviation Administration has confirmed Gustav E. Lundquist as associate administrator for engineering and development. He had been acting administrator. David H. Israel, former No. 2 man of the Defense Communications Planning Group in the Defense Dept. has been named director of the FAA’s Office of Systems Engineering Management. He is a pioneer in the use of computers in air traffic control, the FAA says... NASA has decided that its proposed Orbiting Solar Observatory needs to be bigger, to handle larger experimental packages, and also should have more power and an improved command system. So the space agency has increased its contract with Ball Brothers Research Corp. of Boulder, Colo., some $4-million. The contract now comes to about $10-million. A launching is scheduled for mid-1971... Turbotrain service between Boston and New York will continue for another two years under a $3.8-million contract the Dept. of Transportation has signed with the United Aircraft Corp. Two new cars will be added to each train, and all cars will be modified to cut noise and vibration.
have an oscilloscope that digitally measures: frequency, resistance, current, voltage, and temperature. You still retain all normal scope features such as delaying sweep and dual trace. The CRT display at right is just one of many measurements possible with a scope/DMM/counter combination.

The 7D13 Digital Multimeter has 3 1/2-digit readout. It measures DC voltages to 1000 V with an accuracy of ±0.1%, ±1 count; DC current to 2 A; resistance to 2 MΩ; and temperature from −55°C to +150°C. The 7D13 input can be floated up to 1.5 kV above chassis potential. This allows considerable flexibility in measuring parameters that have a high common-mode voltage. A unique probe is supplied for measuring both voltage and temperature.

The 7D14 Digital Counter is directly gated to 500 MHz and has 8-digit readout. Both 50-Ω and 1-MΩ inputs are provided. Sensitivity is 100 mV P-P (35 mV RMS), about three times better than most counters. The signal connected to the vertical amplifier can also be routed to the 7D14 through the oscilloscope's trigger source switches. All eight 7000-Series vertical amplifier plug-ins (differential comparator, 10-μV differential, current amplifier, etc.) are available as signal conditioners for the counter.

The 7D14 will determine ratios from 0 to 10^4 and totalize from 0 to 10^8. The delayed sweep from the oscilloscope can drive the counter gate. By doing this, signals are displayed on the CRT with the ones being counted intensified.

For complete information on these exciting plug-ins, contact your nearby Tektronix field engineer or write: Tektronix, Inc., P.O. Box 500, Beaverton, Oregon 97005.

Prices of instruments shown: 7504 90-MHz Oscilloscope $2000, 7A12 Dual-Trace Amplifier $700, 7B52 Dual Time Base $900, 7D13 Digital Multimeter $560, 7D14 Digital Counter $1400. The 7D13 and 7D14 are compatible with all five 7000-Series mainframes.

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editorial

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One of the most important assets of a magazine like ELECTRONIC DESIGN is its readership. In keeping with this concern, the editors are turning the editorial page in this issue over to our Circulation Manager.

My primary concern, as Circulation Manager, is to maintain your continued subscription, mailed personally to you.

In the Nov. 8, 1970, issue we initiated a new technique for renewals. A renewal card was bound into each copy, and a special label was supplied that could easily be affixed to the renewal card.

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Nancy Merritt

[Signature]
Good microstrip multipliers don't just happen. A solid step-by-step design approach is needed, and early testing of the subcircuits is advised.

Second of two articles

Designing microstrip frequency multipliers isn't easy. All multipliers, of course, are grossly nonlinear, making their analysis difficult. And the varactor is a two-terminal device, which means that the input and output circuits of varactor multipliers will tend to interact.

These two problems are common to all varactor multipliers. And when the difficulties of designing in microstrip are added to them, the task can become overwhelming if it is not attacked in a methodical fashion. The following 10-step design procedure has been found to be an effective aid:

Step 1. Calculate the diode parameters demanded by the application.

Step 2. Select a varactor diode from Step 1. This selection involves a decision about the use of a packaged device or an unmounted chip.

Step 3. Calculate the circuit parameters for the specified diode, drive level and frequencies. (Burckhardt1 provides an excellent discussion of the calculation of circuit parameters.)

Step 4. Decide whether or not to include an idler in the multiplier.

Step 5. Recalculate the circuit parameters with the idler (if one is used) unless the idler is to be incorporated in the filters.

Step 6. Design the input filter. Breadboard and test it, if possible. Decide on the design of the bias network.

Step 7. Design the output filter. Breadboard and test it, if possible.

Step 8. Breadboard and test a scaled-up version of the complete multiplier by building it on a thin piece of material with a low dielectric constant. Teflon fiberglass 1 32 inch thick is a good choice.

Step 9. Transpose the design to microstrip.

Step 10. Build and test the microstrip multiplier.

To get a feel for some of the practical problems that come up in the design of multipliers, let's go over the design of an actual doubler to 4 GHz.

The chief goals are to minimize the size of the doubler, to be able to handle 2 W of input power at 2 GHz and to have a bandwidth of 1 to 2%.

Because of the power requirement, a shunt-mounted Motorola type 1N5154 is selected. To maximize the power-handling capability of the chip, a hole is made in the substrate material and the diode is mounted directly onto a copper base plate.

From the manufacturer's data sheet, the diode's circuit parameters are calculated to be:

\[ R_{in} = 10 \, \Omega \]
\[ C_{in} = 3.4 \, \text{pF} \left( = -j23.4 \, \Omega \right. \text{ at 2 GHz} \]
\[ R_{out} = 19 \, \Omega \]
\[ C_{out} = 3.4 \, \text{pF} \left( = -j11.7 \, \Omega \right. \text{ at 4 GHz} \]

Diode efficiency = 60%

Alumina (\( \epsilon_r = 9.5 \)) is used as the substrate material, because of its low cost and ready availability. A standard thickness of alumina blanks is 0.025 inch; it will be used in the design.

As the multiplier is only a doubler, there is no need for an idler circuit. These circuits are needed only in higher-order multipliers to deal with the undesired harmonics. (For example, a tripler with an input frequency of \( f_0 \) and an output frequency of \( 3f_0 \) may exhibit enhanced efficiency if an idler at \( 2f_0 \) is included in its design.)

Use resonant-line filters

Since the design goals require only a 1 to 2% bandwidth, resonant-line filters can be used on both the input and output. This type of filter consists of a quarter-wave open-circuited shunt stub located a quarter-wavelength away from the diode. The filter blocks the frequency at which the quarter-wave relationships hold. Since only the lengths of the quarter-wave sections are specified for frequency separation, the impedances can be adjusted to match the diode.

In choosing the impedances, care must be exercised to ensure that the resulting line widths are physically practical. That is, a thick-film process may give repeatable results only with line widths of 0.005 inch or greater. For a given substrate material and thickness, this sets an upper limit on

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1. \( Z_{\text{diode}} \) is rotated an eighth of a wavelength at \( f_0 \) by the line connecting the input filter to the diode. This transforms it into \( Z_0 \), or, equivalently, to \( Y_0 \) which turns out to be 50 \( \Omega \) in parallel with an inductance.

2. A 29-\( \Omega \) line is used at the output to match the diode to 50 \( \Omega \). The line is extended beyond the quarter-wave point (at 2\( f_0 \)) to add some inductive reactance, which cancels the diode series capacitive reactance.

\( Z_0 \). Similarly, the lower limit on \( Z_0 \) is set by the difficulty in connecting low-\( Z_0 \) (wide) and high-\( Z_0 \) (narrow) lines.

An important factor in these calculations is the lead inductance from the microstrip to the chip. The wire bond (0.1 inch) has been calculated to have an inductance of approximately 0.2 nH.

Since the diode impedances are now known, let's match it to the 50-ohm source and load. Assuming no interaction between input and output, the diode chip impedances with lead inductance are:

\[
Z_{\text{in}} = 10 - j23.4 + j2.5 \, \Omega \\
= 10 - j20.9 \, \Omega
\]

and

\[
Z_{\text{out}} = 19 - j11.7 + j5.0 \, \Omega \\
= 19 - j6.7 \, \Omega.
\]

Smith charts ease the job

If the diode input impedance is normalized to 52 ohms, it becomes \( Z_{\text{in norm}} = 0.192 - j0.402 \). To design the input filter, this impedance is plotted on a Smith chart (Fig. 1) and rotated a quarter-
3. Rotating an open circuit, Point A, by 0.155λ converts it to j20 Ω, Point B, on a 29-Ω line. On a 50-Ω line, j20 Ω is at Point C so that only 0.06λ of additional rotation is needed to transform the open into a short.

wavelength at 2f₀, where f₀ is the input frequency. This corresponds to an eighth of a wavelength at f₀. If the transformed impedance, Zᵢ, is converted to an admittance, Yᵢ, it appears to be a shunt 50-ohm resistance in parallel with an inductive reactance of 25.6 ohms.

To eliminate this reactance, it can be resonated with the quarter-wave (at 2f₀) stub. Since the length of the stub is fixed, its impedance must be chosen to provide a capacitive reactance of 25.6 ohms at f₀.

The impedance of an open stub less than a quarter-wave long is given by

\[ Z = -jZ₀ \cot \left( \frac{2\pi f}{\lambda} \right) \]  

where \( Z₀ \) is the characteristic impedance of the stub and \( f \) is its length. The stub is a quarter-wave long at 2f₀; therefore, at f₀, \( f = \lambda / 8 \). Plugging into Eq. 1 yields

\[ Z = -jZ₀ \cot \left( \frac{\pi}{4} \right) \]

or

\[ Z = Z₀ \tan \pi 4 = Z = 25.6 \Omega. \]

Now, if desired, a stub that is a quarter-wave long at 3f₀ may be added at the same point to

4. The bias is introduced before the input filter to avoid the formation of a low-frequency idler. Note that this complete doubler design includes a stub to block the third harmonic at the input.
block the third harmonic. If this stub is added, a new calculation of impedances for both stubs will have to be performed.

For the output filter, a slightly different approach is used. The same idea of putting a quarter-wave stub a quarter-wave away from the diode is used to block the input frequency, \( f_a \). But instead of using a 50-ohm line between the diode and the stub, and adjusting the stub impedance to match the diode (at \( 2f_a \)), a line will be selected with an impedance that is the geometric mean between the output resistance of the diode and the desired impedance (50 ohms). If the diode's output impedance (at \( 2f_a \)) were a pure resistance, then the geometric-mean matching section would be a quarter-wave long at \( 2f_a \). However, the diode has a series capacitive reactance, making it necessary to increase the length of the matching section.

To find out how long to make the matching section, we again turn to the Smith chart (Fig. 2). The geometric mean between 19 and 50 ohms is 31 ohms. Using the chart and a normalizing impedance of 31 ohms, it becomes evident that the match cannot be made simply with one length of line. However, switching from 31 to 29 ohms makes the job quite simple.

The normalized diode output impedance is \( 0.66 - j0.23 \) ohms, which when rotated by \( 0.31\lambda \), matches the diode to 50 ohms at \( 2f_a \). The quarter-wave (at \( f_a \)) stub spaced approximately a quarter-wave (at \( f_a \)) from the diode does not affect the match at \( 2f_a \), because it looks like an open circuit at that frequency.

The quarter-wave (at \( f_a \)) stub is not placed exactly a quarter wavelength from the diode because the output line is not uniform. The line has a 29-ohm section that is 0.155\( \lambda \) long (0.31\( \lambda \) at \( 2f_a \), equals 0.155\( \lambda \) at \( f_a \)) and the remainder has a characteristic impedance of 50 ohms. To see where to place the stub, we recall that we want it to look like an open circuit at the diode.

Rotating an open circuit by 0.155\( \lambda \) transforms it to \( 29 \times j0.69 = j20 \) ohms on a 29-ohm line (Point B in Fig. 3). Switching from a 29-ohm line to a 50-ohm line moves the impedance from B to C, which means that only 0.06\( \lambda \) of additional rotation is needed to transform the open circuit into a short circuit. Hence, the \( \lambda/4 \) stub should be located \( 0.155\lambda + 0.06\lambda = 0.215\lambda \) from the diode.

Note that no filter is needed to block the third harmonic, because the quarter-wave stub at \( f_a \) blocks all of the old harmonics.

**Self-bias is used**

The varactor is biased by adding a resistor between the input and ground (Fig. 4). When the input drive signal goes negative, the resistor provides a path through which the input coupling capacitor, \( C_v \), charges up to a negative potential. The bias resistor is variable, to allow adjustment of the bias voltage with changes in input drive level.

As suggested in the first article in this series, the bias is applied through the input filter through a quarter-wavelength capacitively shorted line. This lowers the probability that the bias circuit will cause problems by creating a low-frequency idler. An even better technique is to use a resistor instead of a quarter-wavelength shorted line; this is more effective at suppressing idlers over a wide bandwidth.

The doubler of Fig. 4 was actually constructed on a 1-inch \( \times \) 1-inch alumina substrate (Fig. 5), and driven to a power output of 1 W. The doubler had an over-all efficiency of 45\%, a maximum VSWR of 1.5, and a 1-dB bandwidth in excess of 5\%.

The difference between the calculated diode efficiency (60\%) and the measured over-all efficiency (45\%) was caused by losses in the input and output circuits.

Note that the resonant line filter used at the output has been bent to save space—a good technique to use when small size is important.

---

**References**


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After helping a jillion feet of paper tape wind and unwind its way through communications systems everywhere, Teletype announces the addition of magnetic tape data terminals.

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Divide frequencies by any integer.
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The division of clock frequencies by any integral divisor can be accomplished by selectively controlling the input signals to a common pulse-gated ripple counter. The number of components is reduced in comparison to other methods of frequency division because only one monostable-gate generator is needed. This advantage is most apparent for large input/output frequency ratios that would otherwise require many external gates.

Either TTL or RTL logic can be used for the circuit, and the method is easily adaptable to MSI or conventional manufacturing. For example, 10 flip-flops and four dual gates, either deposited on one MSI chip or assembled from ICs onto a circuit card, can provide a low-cost, adjustable divider to cover the entire audio range from one clock frequency.

Timing determines the divisor

The basic operating principle of the divider is the control of gating of flip-flops in a ripple counter.

Counts that are powers of two are obtained directly without any external gating. A two-stage counter divides by four; a three-stage by eight; an n-stage by $2^n$. When a four-stage counter is used, for example, every 16th clock pulse causes one output pulse to appear.

Now, if the first flip-flop stage of the counter does not respond to one clock pulse in 16, it will take 17 clock pulses for the output flip-flop to flip. If two clock impulses in 16 are inhibited in the first flip-flop, 18 clock pulses will be required to make the fourth flip-flop respond. In the first case, the clock frequency, $f_c$, is divided by 17; in the second case it is divided by 18.

Let $f_o$ be the output frequency, $n$ be the number of flip-flop stages in the counter and $M$ be the number of clock pulses inhibited. Then

$$f_o = f_c / (2^n + M)$$

Or $f_o = 2^n + M$

How are the clock pulses inhibited at the first flip-flop? If each output pulse from the last flip-flop enables a monostable-gate-pulse generator, the monostable output can be applied to the inhibit input of the first flip-flop. By adjusting the length of the monostable pulse, the first flip-flop can be made to ignore one, two or more clock pulses each time the output stage goes high.

The circuit using TTL flip-flops that performs in this way is shown in Fig. 1. The duration of the inhibiting monostable-gate pulse is determined by the value of $C$ in Fig. 1. The resistance of the RC network is the output of the NAND gate, typically 1 kΩ. If $C$ is made variable, the duration of the inhibit period is also variable, and the value of $M$ is adjustable.

The frequency ratio of a counter can be made to be less than $2^n$ by using a slightly different technique. In this case, the ratio is $2^n - M$, and $M$ extra pulses must be inserted to make the first flip-flop “think” that more, rather than fewer, clock pulses have appeared. Using a four-stage counter as an example, an extra pulse at the first flip-flop, triggered by the output pulse, results in an output response for only 15 clock pulses—or division by 15.

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1. The division ratio of a four-bit ripple counter will increase to more than 1/16 if a variable width inhibit pulse (controlled by RC) is applied.
However, a difficulty arises in this approach: making \( M \) greater than one is impractical because there is not enough time between clock pulses to insert more than one extra pulse at the first flip-flop stage.

The solution to the problem is to use only one short pulse, but to insert it at different stages. As previously described, one extra pulse at the first stage \((M=1)\) reduces the frequency division ratio, \( f_d / f_a \), by one. If the extra pulse is inserted only at the second stage, the ratio would be reduced by two \((M=2)\). If inserted at both the first and second stages, the ratio is reduced by one plus two \((M=3)\). Thus \( M \) can be made to have any negative value up to \( 2^n \) by inserting the extra pulse at the proper stage or stages. Note that the extra pulse must occur between clock pulses.

In order to generate the additional gating pulse, a monostable-gate-pulse generator is again used. However, the pulse generated need not be adjustable in duration, and it must be shorter than a clock pulse.

A convenient means of directing the monostable pulse to the proper flip-flop stage is to use an octal switch (see box). A nine-stage divider using octal switches is shown in Fig. 2. This circuit, with RTL flip-flops, can divide the clock frequency by any integer from one to 512 by proper setting of the octal switches.

The stability of the circuit is primarily dependent on the RC-time constant of the monostable-gate generator. This stability can be very good in circuits with negative values of \( M \), because fixed values of \( R \) and \( C \) are used. The stability is not as good in circuits with positive \( M \), particularly as \( M \) becomes larger, because adjustable values of \( C \) are used.

**What is an octal switch?**

An octal-coded switch is a three-gang selector switch with two, four and eight segments on each gang as shown in the diagram. Its output, at the terminals labeled 1, 2 and 4, constitutes a single parallel octal word or byte.

If this switch is to be used in a frequency divider in a subtract mode, the input terminal is connected to the gate pulse generator output, and the terminals 1, 2 and 4 are connected to the corresponding \( C_d \) or \( S_d \) terminals of the ripple counter. This first word or 8° switch is used with the first three stages of the counter. A second word or 8¹ switch would be used with the 8, 16 and 32 stages, and so on.

---

2. **An RTL ripple counter with nine stages** is similar in operation to the TTL circuit of Fig. 1. By selectively connecting the gate pulse to the \( C_d \) inputs through three octal switches, division ratios from one to 512 can easily be chosen by setting the switches to the appropriate values. Adjustable pulse widths are not required.
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Selecting preamps for lowest noise can be difficult if the candidates aren’t described the same way. Here’s how to make the most commonly needed conversions.

A problem forever confronting the engineer in his quest for improved signal-to-noise (S/N) ratio is that of comparing preamplifiers offered for sale. One manufacturer describes the noise performance of his amplifier by providing a plot of equivalent noise-voltage density \(e_n\) and equivalent noise-current density \(i_n\) as functions of frequency. Another prefers to provide a plot of \(R_e\) and \(R_i\) as functions of frequency, where \(R_e\) is the Johnson noise equivalent of the amplifier’s input short-circuit noise-voltage generator and \(R_i\) is the Johnson noise equivalent of the amplifier’s input open-circuit noise-current generator. A third manufacturer provides full noise-figure contours for his amplifier. As a result, the researcher is all too frequently faced with the problem of comparing apples with oranges with pears.

**Noise-figure contours are best**

It is generally conceded that noise-figure contours are the best means of specifying the noise characteristics of an amplifier. Each contour is the locus of points of constant noise figure on a graph whose ordinate is the source resistance and whose abscissa is the frequency of operation. The complete set of contours (Fig. 1) allows the experimenter to quickly determine how much the amplifier will degrade the input S/N ratio for any specified frequency and source resistance.

And, fortunately, the formulas relating noise figure to \(R_e\) and \(R_i\), and \(e_n\) and \(i_n\), are simple. Noise-figure contours can easily be generated when the noise data is in terms of \(R_e\) and \(R_i\), or \(e_n\) and \(i_n\). If only a few points are of interest, these same formulas can be used to compute the noise figure at these points, and the evaluation can be quickly completed without plotting the contours.

Assume that a prospective preamplifier buyer is holding two specification sheets—one showing a set of noise-figure contours for amplifier A (Fig. 1), and the other a plot for \(R_e\) and \(R_i\) as functions of frequency for amplifier B (Fig. 2). To convert the \(R_e\) and \(R_i\) curves into NF contours, one can use the formula

\[
NF \ (dB) = 10 \log_{10} [1 + (R_e / R_s) + (R_i / R_f)]
\]

where \(R_s\) is the source resistance from which one intends to operate.

A good way to begin comparing amplifiers is to compute the minimum noise figure, \(NF_{\text{min}}\), for a few values of \(R_e\) and \(R_i\), to see if the two amplifiers are in the same ballpark. \(NF_{\text{min}}\) is given by

\[
NF_{\text{min}} \ (dB) = 10 \log_{10} [1 + 2\sqrt{R_e / R_i}] \quad (2)
\]

For example, amplifier B will have a minimum noise figure at 10 kHz of approximately 0.03 dB, indicating immediately that the amplifier does indeed have excellent noise performance. By checking a few other points, it becomes clear that amplifier B is quieter than amplifier A. The whole story is told by plotting the noise figure contours for amplifier B and comparing them with those provided for amplifier A.

**Plotting the contours**

The first step in plotting the contours is to solve Eq. 1 for \(R_s\). The result is

\[
R_s = 1 / 2 [R_e (10^{N/10} - 1)]^{1/2} + \sqrt{(R_e - 10^{N/10} R_i)^2 - 4R_e R_f}] \quad (3)
\]

Next, decide on the contours to be plotted. The same contours as are shown for amplifier A are the logical choice. What remains is simply to insert a value of noise figure, along with \(R_e\) and \(R_i\) at one frequency, into Eq. 3, and solve for the two values of \(R_s\) that yield the desired noise figure.

By repeating the process enough times with different values of \(R_e\) and \(R_i\), enough values of \(R_s\) will soon be accumulated to enable one contour to be plotted. The entire process is then repeated for as many contours as desired. A logical choice of values for \(R_e\) and \(R_i\) are those at 1-2-5 frequency multiples. These values will result in three sets of values for \(R_s\) per frequency decade, with approximately equal spacing between sets when plotted on log log paper. In general, one should begin with the smallest contour of interest and work outward.

---

Admittedly this is a lot of work, and Eq. 3 gets messy with the values typically taken by R, and R,. Fortunately, small, reasonably powerful desktop computers abound that can handle Eq. 3 with ease, particularly if \(10^{\text{NF}/10}\) is evaluated in advance.

An interesting characteristic of the noise-figure contours is that for a low-noise amplifier having \(R, \gg R,\) the 3-dB contour contains \(R,\) and \(R,\) to a good approximation, with the upper curve of the contour representing \(R,\) and the lower representing \(R,\). Over the region where the upper and lower curves are widely separated, the approximation is quite close. At the ends, where the curves converge and close, the approximation is inaccurate.

This closeness of the 3-dB contour to the \(R,\) and \(R,\) curves gives us another method for quickly comparing amplifiers. By estimating \(R,\) and \(R,\) from the 3-dB contour, one obtains noise-performance data for both amplifiers in the same terms. However, it is frequently not obvious from the \(R,\) and \(R,\) plots which of the two amplifiers will be quietest at a particular combination of frequency and source resistance, particularly when the regions bounded by the two sets of \(R,\) and \(R,\) are not centered on the same point. To estimate the relative noise performance at a given point of interest, Eq. 1 must be used.

Converting from \(e_n\) and \(i_n\)

Except for one additional step—that of converting \(e_n\) and \(i_n\) to \(R,\) and \(R,\)—the procedure is exactly the same as before. The conversion equations are

\[ R, = e_n^2 / 4kT \]  \hspace{1cm} (4)

and

\[ R, = 4kT / i_n^2 \]  \hspace{1cm} (5)

where \(k\) is Boltzmann’s constant, \(T\) is the absolute temperature, \(e_n\) is the equivalent input noise-voltage density in \(\text{V Hz}^{1/2}\) and \(i_n\) is the equivalent input noise-current density in \(\text{A Hz}^{1/2}\).

For convenience, one should again make the conversions at frequency multiples of 1-2-5 to obtain three approximately equal spaced points per decade on log/log paper. Once the conversions are completed, the values of \(R,\) and \(R,\) can be converted to noise-figure contours as directed earlier.

Another way of comparing amplifiers is to compute the total noise referred to the input \((E_{n1})\) given any of the three ways of specifying noise performance. This will often prove the quickest and easiest way of estimating the relative noise performance of different amplifiers at the intended operating frequency and source resistance. \(E_{n1}\) is expressed as a function of NF and as a function of \(e_n\) and \(i_n\) in the two formulas that follow:

\[ E_{n1} = \sqrt{4kTR,} \times 10^{\text{NF}/10} \hspace{1cm} \text{Vrms/Hz}^{1/2}, \hspace{1cm} (6) \]

\[ E_{n1} = \sqrt{4kTR, + e_n^2 + (i_n R,)^2} \hspace{1cm} \text{Vrms/Hz}^{1/2} \hspace{1cm} (7) \]

Note that Eq. 7 can be used to convert \(R,\) and \(R,\) to \(E_{n1}\) by first using Eqs. 4 and 5 to convert \(R,\) and \(R,\) to \(e_n\) and \(i_n\). ■

Test Your Retention

Here are questions based on the main points of this article. Their purpose is to help you make sure you have not overlooked any important ideas. You’ll find the answers in the article.

1. Why are noise-figure contours the preferred way of specifying an amplifier’s noise figure?
2. Do NF contours contain any information not contained in plots of \(R,\) and \(R,\) or \(e_n\) and \(i_n\)?
If your problem is in-circuit testing of transistorized and integrated circuits . . . Solve it with Triplett's 601

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It has 7 low-power resistance ranges that apply only 75 mV to the device under test . . . does not activate or damage solid-state component . . . full-scale DC measurements down to 100 mV and 10 μA and AC as low as 10 mV and 10 μA, it's obvious the Model 601 was designed for in-circuit testing.

Add such features as 10 megohm input impedance on AC and 11 megohm input resistance on DC, voltage readings to 2% DC and 3% AC (current: 3% DC and 4% AC), separate range-selection and function-selection switches, and a simplified dial on which all 53 ranges are read on only 4 scales, and it's equally obvious that here's a V-O-M that has what you need to do the job better, faster and more easily. See the capable Model 601 — priced at $166 — at your local Triplett distributor. For more information, or for a free demonstration, call him or your Triplett sales representative right away. Triplett Corporation, Bluffton, Ohio 45817.

In the capable Model 601 —
Stabilize a constant-current supply with an op amp

Constant-current sources usually perform adequately, but they can be subject to errors from several sources. Some of the errors are variations in power-supply level; variations in transistor parameters due to changes in temperature or current; and variations in the reference voltage of a zener diode due to current changes. The constant-current circuit shown in the diagram uses an IC op amp that overcomes all of the errors listed, at the cost of only a few additional components.

If \( R_i=R_s \) and \( R_i=R_s/(R_t+R_s) \) in the circuit, the output current, \( I_o \), is given by:

\[
I_o = \frac{R_iE_i}{R_iR_s}
\]

This result indicates that the load current is independent of the supply voltage and the load resistance. It depends only on the resistances of three resistors and the reference voltage of a zener diode. In addition, the zener diode is isolated from the load so that its current does not change. Its reference level is thus constant, and as a result \( I_o \) is stable.


Vote for 311

Voltage-to-frequency converter provides linearity of 0.4%

A simple low-cost circuit can be used as either a sawtooth generator or as a voltage-to-frequency converter with a system linearity of better than 0.4% over a two-decade frequency range. The temperature stability of the circuit is almost exactly that of the unijunction used.

Due to the negative input voltage \( V_i \), the current through \( R_i \) must be equal to the current in the feedback capacitor \( C \). This assumes that the amplifier (Fairchild’s \( \mu A741 \)) has sufficient gain to make its input current negligible.

For a constant \( V_i \), the voltage at point A increases linearly with a slope of \( V_i/R_sC \) volts per second. \( R_s \) is chosen so that its current is always greater than the current through \( R_i \). Excess current, therefore, flows to the amplifier through \( D_2 \). When point A reaches the triggering potential of the unijunction, point A is shorted to ground via \( R_i \), and \( C \) discharges through \( D_2 \).

During this discharge period, the output of the amplifier will reach its positive saturation potential. After the rapid discharge of \( C \), the cycle is repeated. For frequencies greater than a few kilohertz, it is better to use an amplifier with a higher slew rate, like Fairchild’s \( \mu A709 \).

G. Pranzo Zaccaria, Engineer of European Space Research and Technology Centre, 40 Estec, Noordwijk, Holland.

Vote for 312
Inexpensive isolator circuit solves interface problems

A low-cost circuit can replace expensive line-driver receiver elements at a parts cost of less than $5 per information line, while alleviating cumbersome intersystem wiring. It can solve problems like interfacing digital information between systems that operate from isolated power sources, and interfacing in the presence of large ground potential differences.

The major components would be three DIP quad transformers (Pulse Engineering) and two hex inverters (Signetics' 8H90). To make full use of all device packages, 12 stages represent the least common denominator.

Data transitions differentiated by \( R \) and \( C \) commence transformer action, which is sustained by positive feedback through \( A_1 \) and \( A_2 \), the active pull-up inverters. As the transformer approaches and finally reaches saturation, the positive feedback is maintained through the low dc resistance of its secondary. Resistor \( R_s \) is necessary as a current path for the transformer, since \( A_1 \) is a TTL element and cannot sink current flowing out of the transformer secondary during positive half cycles.

This circuit has been operated with ±20% supply voltage variations over a range of −55 to +85°C. Performance speed can vary from 1 to 5 × 10³ pulses per second. Propagation delay is less than 50 ns.

Bud Norris, Design Engineer, Leach Corp., Controls Div., P. O. Box 820, 1499 Huntington Dr., South Pasadena, Calif. 91030.

VOTE FOR 313

Temperature-compensated dc restorer offers variable clamping

An inexpensive monolithic transistor array can be used as a low-drift dc restorer with a variable clamping level and good interfacing characteristics. The circuit, RCA’s type CA3018, uses \( Q_s \) and \( Q_t \) as a temperature-compensated voltage source and \( Q_r \) as a clamping transistor.

With no input, \( Q_s \) is saturated via \( R_t \) and its emitter potential closely approaches that of \( Q_r \). This is the high-output state, whose level is determined by the emitter of diode-connected \( Q_r \).

When a signal is applied, \( Q_r \)'s base-collector junction rectifies positive signal peaks, while the base of \( Q_s \) is clamped one diode drop more positive than \( Q_r \)'s emitter. Since \( Q_s \) is saturated, its two junction drops cancel, and the positive signal peak is referenced to \( Q_r \)'s emitter.

During the negative portion of the input, \( Q_s \) comes out of saturation and passes the signal as a normal emitter-follower. \( Q_s \) can be used as an additional emitter-follower as shown.

Walter Jung, Senior Engineer, Control Concepts Corp., Rockville, Md.

VOTE FOR 314
new products

DPM uses LED display and custom parts for reliability

Hewlett-Packard Co., Loveland Instrument Div., P. O. Box 301, Loveland, Colo. Phone: (303) 667-5000. P&A: $225 in quantity; stock by May.

Using light-emitting diodes in a 4 by 7-dot matrix to display 0 to 1.000 V, the model 3431A digital panel meter with custom components for increased reliability dissipates only 5 W of power and measures 1.7-in. high by 3.5-in. wide and just 2.9-in. deep.

This single-range 11-oz DPM features an accuracy of ±0.1% of reading ±1 digit and includes 100% overranging capability with the last three displayed digits blanking for overload measurements.

The 3431A is characterized by many distinctive features such as full-scale and 10% of full-scale auto-ranging flags, a 55-ms response time from zero to full scale and programmable sampling rate of 0, 1, 2, 4, 8 and 15 readings/s.

Other features include automatic and manual polarity and remote polarity selection. Hold, read, trigger and decimal locating functions are all also remotely controlled.

An integral power supply can be operated from either 115 V ac over 50 to 400 Hz or optionally from a 5 to 15-V dc source for portable applications.

Other characteristics include an input impedance of 10 MΩ, low bias current of 30 nA and BCD output (serial-digit parallel-bit). Common-mode rejection ratio is 80 dB and the operating temperature range is from 0 to 60°C.

A host of options are available with the 3431A DPM. These include 1 to 3 remotely controlled LED annunciators and remote display for panel-space miniaturization applications.

Other options include full-scale voltage ranges of 10.00 and 100.0 V and the isolation of digital and analog grounds by 500 V.

Later models will have additional options, including parallel-character BCD output for printers and data loggers, a 230-V ac power supply, and other dc power supplies.

By designing the 3431A with custom IC components total parts count is significantly reduced, thereby enhancing reliability. The meter's metal front panel pops out to unveil the mounting hardware and service adjustments. The thin panel also allows the user to easily paint or silk-screen the panel to his own taste.

CIRCLE NO. 250

Plug-in display together with plug-in PC boards (left) simplify servicing in the new 3-1/2 digit LED DPM (right).

31-range multimeter retails at only $89

Epic, Inc., 150 Nassau St., New York, N.Y. Phone: (212) 849-2470. Price: $89.

The Elavi 6 is a low-cost portable multimeter housing 31 measurement ranges for only $89. It checks 150 mV to 600 V dc at 1.5% accuracy, 6 to 600 V ac at 2.5% accuracy and resistance in two ranges of 1 and 10 MΩ. It also measures dc current from 0.3 to 1500 mA and ac current from 0.3 to 6000 mA. Internal resistance is 33 kΩ/V and 10 kΩ/V for dc and ac, respectively.

CIRCLE NO. 251

Counter for $625 reads out to 50 MHz


Covering the frequency range from dc to 50 MHz, a new five-digit universal counter sells for only $625. Model 1192-B offers an internal room-temperature crystal oscillator and an over-all stability of better than 2 ppm per month. Its sensitivity is 10 mV up to 20 MHz, 20 mV to 35 MHz, and 30 mV to 50 MHz. Measurement parameters include frequency, period and time interval.

CIRCLE NO. 252
MIC YIG filter is 2.1-in. long


Able to use YIG-tuning techniques and still be compatible with microwave integrated circuits, a new YIG-tuned dual-stage filter (two separate two-stage filters within the same magnetic structure) measures a mere 0.75 in. in diameter and only 2.1 in. in length.

Designed for use in the S band from 2 to 4 GHz, the new YIG-tuned filter is adaptable to two, three and four-stage filter configurations in addition to the two-stage designs mentioned. It can also be used as the magnetic structure for YIG-tuned oscillators.

An important characteristic of the new YIG-tuned filter is its compatibility with stripline packaging. It uses stripline output tabs for direct interconnections to stripline or microstrip assemblies. If the user desires, small coaxial type connectors can also be fitted on the filter.

Despite the fact the YIG-tuning techniques require large magnetic fields and consequently large currents, the filter's small size and tabular or solenoidal shape results in an optimum configuration for low tuning power. It dissipates only 0.45 W at 4 GHz compared with conventional cubical filters which dissipate 0.6 W.

This reduction in power reduces the heat which must be dissipated within the microwave integrated circuit package, thus minimizing heat dissipation problems.

An additional advantage is the reduction in the filter's coil inductance to about one-half that of conventional filters thus reducing filter/driver interface problems for fast sweep-rate devices.

2-in.$^3$ mixer-preamp covers 1 to 18 GHz

RHG Electronics Laboratory, Inc., 94 Milbar Blvd., Farmingdale, N. Y. Phone: (516) 694-3100.
P&A: $850; 45 days.

A new 2-in.$^3$ hybrid IC mixer preamplifier includes a mixer with multi-octave rf coverage for the frequency range of 1 to 18 GHz. Model DM1-18/10HH has instantaneous i-f coverage from 10 to 200 MHz. Noise figure ranges over 7.5 to 11 dB and VSWR is 2:1. Isolation is greater than 20 dB and output capability is greater than 0 dBm.

2 to 18-GHz antenna has a VSWR of 2

American Nucleonics Corp., 6036 Variel Ave., Woodland Hills, Calif. Phone: (213) 347-4500.

A new broadband cavity-backed spiral antenna, model AM-277, performs over a frequency range of 2 to 18 GHz with a VSWR of less than 2.0 and a VSWR of 1.0 over the frequency range of 2 to 16 GHz. Gain is 2 dB over the frequency range of 7 to 16 GHz. Units are available with right-hand circular polarization with several optional connector locations.

Stripline varactor diode rates at 90 GHz


A new GaAs varactor diode is available with three figure-of-merit ranges from 40 to 90 GHz. The device, designated XMD38A, is a diffused mesa diode mounted in an LID package in a square configuration for stripline applications. Package dimensions are 0.15 by 0.045 by 0.024 in.

1-GHz crystal source supplies 2 W cw

Trak Microwave Corp., 4726 Eisenhower Blvd., Tampa, Fla. Phone: (813) 884-1411.

A new crystal-controlled 1-GHz oscillator supplies 2 W for cw and pulsed operation. Stability of the 5038-1100 unit is ±0.001% and fundamental frequency is 125 MHz. The hybrid thin-film and lumped constant device operates from ±12 V dc and measures 1 by 1.5 by 6 in.

4 to 18-GHz amplifiers have 4-GHz bandwidths

RCA Electronic Components, 415 S. 5th St., Harrison, N. J. Phone: (201) 485-3900.

New cw linear amplifiers for use in C, X and Ku-band frequencies (4 to 18 GHz) provide instantaneous bandwidths of 4 GHz or more. They make use of the negative-resistance characteristic of transferred-electron devices to obtain stable broadband amplification.

750-foot-lambert LED costs just $1.50

Litronics, Inc., 10440 N. Tantau Ave., Cupertino, Calif. Phone: (408) 257-7910. P&A: $1.50; stock.

The Red-Lit 50 is a visible-red LED with a brightness of 750 foot-lamberts and current drain of 20 mA selling for only $1.50. The unit uses an epoxy package 80-mils wide that is elongated in the direction of the leads. Viewing angle is 45 degrees.
DATA PROCESSING

2000-word wire ROM accesses in 150 ns


Utilizing nondestructive readout wires, the FS-100A read-only wire memory allows repeated readouts with an access time of 150 ns and a cycle time of 200 ns. It has a memory capacity from 1-k words at 72 bits each up to 2-k words at 72 bits each. Despite its construction as a fixed memory for readouts only, it is designed to enable swift rewriting electronically.

CIRCLE NO. 260

New low cost LED Vactrol photon isolator

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as low as $4.20 each in 1,000 quantities

- all solid state
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- unlimited life—no filaments
- ideal for environments where shock and vibration are a problem
- applications include photochoppers, linear isolators, noiseless switching, SCR and triac turn-on, audio level controls, etc.

<table>
<thead>
<tr>
<th>Part Number</th>
<th>LED Current (ma) (1.65V typ.)</th>
<th>Max. Cell Resistance</th>
<th>Typical Rise Time (ms)*</th>
<th>Decay</th>
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<tr>
<td>VTL2C1</td>
<td>40</td>
<td>10 KΩ</td>
<td>.5</td>
<td>3.5 ms **</td>
</tr>
<tr>
<td>VTL2C2</td>
<td>40</td>
<td>500 Ω</td>
<td>3.5</td>
<td>500 ms †</td>
</tr>
<tr>
<td>VTL2C3</td>
<td>40</td>
<td>2 KΩ</td>
<td>2.5</td>
<td>35 ms †</td>
</tr>
<tr>
<td>VTL2C4</td>
<td>40</td>
<td>100 Ω</td>
<td>6.0</td>
<td>1.5 sec †</td>
</tr>
</tbody>
</table>

* To 63% conductance   ** To 1 meg   † To 100 KΩ

Write for Bulletin VTL2C. Also a complete line of neon and incandescent Vactrol photon isolators.

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CIRCLE NO. 261

Electronic Design 3, February 4, 1971
IC sense amplifier reads 3-mV signals

Signetics Corp., 811 E. Arques Ave., Sunnyvale, Calif. Phone: (408) 739-7700.

NE528B four-channel monolithic sense amplifier for plated-wire memories can read 3-mV signals and translate them to TTL levels. Propagation delay is 20 ns, bias current is 1.5 μA and input resistance is 2 kΩ. Input channels are selected by means of three TTL inputs, one of which is capable of totally disabling the input circuits. The amplifier can be wired OR'd.

CIRCLE NO. 262

Quad peripheral drivers handle 1 A and 180 V


Three new hybrid quad peripheral drivers interface transistors with up to 1 A of current and 180 V of breakdown voltage. The HIC040 drives four 2N3725 transistors with collector current per transistor of 1 A at a 50-V breakdown. The HIC067 has a drive capability of 150 mA and a 40-V breakdown. The HIC068 handles 100 mA at a 180-V breakdown.

CIRCLE NO. 264

2560-bit 500-ns ROM costs only $29

Hughes Aircraft Co., MOS Div., 500 Superior Ave., Newport Beach, Calif. Phone: (714) 548-0671. P&A: $29; 45 days.

The IROM8560 is a monolithic ion-implanted 2560-bit read-only memory featuring an access time of 500 ns and costing only $29 (100-lot quantities). It is organized as 256 words at 10 bits per word and has address input lines decoded on the chip.

CIRCLE NO. 266

1024-bit MOS RAM accesses in 400 ns

Mostek Corp., 1400 Uphield Dr., Carrollton, Tex. Phone: (214) 242-1491. Price: $75.

Using ion-implantation techniques, the MK4006P 1024-bit dynamic MOS random-access memory features an access time of 400 ns, and a write cycle time of 650 ns. It is DTL/TTL compatible and is available in a 16-pin package. Decoding is provided on the chip.

CIRCLE NO. 267

200-V bridges carry up to 3 A


The new BCS300 series of controlled rectifier bridges include 50, 100 and 200-V rms types with output currents to 3 A. Conduction for each rectifier element is through 180 degrees. Packaging is in 1 by 0.5-in. cases.

CIRCLE NO. 268

Dual-gate MOSFET lowers cross modulation

Fairchild Semiconductor, 464 Ellis St., Mountain View, Calif. Phone: (415) 928-3583. P&A: $1.10; stock.

A new low-cost dual-gate MOSFET has cross modulation performance close to that of a vacuum tube for rf applications up to 400 MHz. The FT0601 is designed to reject third-order harmonics.

CIRCLE NO. 269
Solid-state 1-A relay maximizes I/O isolation

Teledyne Relays, 3155 W. El Segundo Blvd., Hawthorne, Calif. Phone: (213) 679-2205. P&A: $50 (803); 12 wks.

Achieving total input-to-output isolation with photon coupling, a new 1-A spdt solid-state DIP flat-pack relay with automatic current limiting for short-circuit protection requires a current control of only 2 mA and operates over the temperature range of -55 to +125°C without the use of any heatsinks.

Series 801 relay eliminates two shortcomings found in previous solid-state relays: inefficiency due to large required control currents and sensitivity due to temperature changes.

Two voltage control ranges of 2.7 to 6 V (5-V digital-logic compatible) and 18 to 36 V are available. Each relay has two input pins for both voltage ranges so that the user need only select the desired input.

The 801 relay features a 5-µs risetime and a typical contact-voltage drop of 0.25 V at 1 A. It achieves its 1-A current rating throughout the entire 4 to 40-V range. Snap-action and 2:1 hysteresis features are built in.

Current is automatically limited to 3 A when the relay’s output load is short-circuited for up to 3 s. This current-limiting capability forms the basis of the 801’s use in a new power controller designated the series 803.

Under no-overload conditions, the 803 power controller (which contains an 801 relay) is identical in performance to the 801 except that it is a spst switch. When overload or short-circuit conditions in the 801 relay persist for more than 3 s, the 803 controller takes over and trips the 801 relay thereby de-energizing the load.

Ferrite toroids cut TC to 0.2%


A new series of ferrite toroids features temperature coefficients as low as ±0.2 to ±0.5% over the temperature interval of -40 to +85°C. The toroids are in sizes of 0.18 in. in outside dia by 0.09 in. in inside dia by 0.05-in. high. Magnetic permeability measured at 5 MHz may be controlled in the range of 20 to 100. Q values at 2 to 12 MHz are from 50 to 75.

Small chip resistors pack in 10^11-Ω values

Eltic Instruments, Inc., Box 46, Lancaster, N. Y. Phone: (716) 683-8421. P&A: $2 (1000-piece lots); 1 wk.

Available in 0.05 by 0.1 by 0.12-in. chips, model 110 glass resistors span a range of 10^8 to 10^11 Ω. Their glass and gold terminals are fired onto the top surface of a ceramic chip. Standard values of 10^6, 10^7, 10^8 and 10^9 Ω have a tolerance of ±25%. Operating temperature is -270 to +200°C and temperature coefficient of resistance is -0.3%/°C.
COMPONENTS

Plastic dpdt relay accepts 1 ampere

Deutch Relay Div., 65 Daly Rd., E. Northport, N. Y. Phone: (516) 266-1600.

Series 3115 dpdt low-cost 1/6-crystal-size industrial relay in a plastic enclosure can handle 1 A of current. The series consists of three models which feature a balanced-armature design to withstand large shock and vibration forces. Operating temperature range is from -25 to +85°C. Pick-up sensitivity is 200 mW at +25°C. Coil resistance ranges from 65 to 1350 Ω.

CIRCLE NO. 273

Preset timing module spans 50 ms to 60 s

Hi-G, Inc., Electronic Products Div., Spring St. & Route 75, Windsor Locks, Conn. Phone: (203) 623-2481.

Series 6100 timing module provides factory preset time delays from 50 ms to 60 s in a DIP package measuring 3/8 by 1/2 by 7/8 in. Application of power to its input initiates the timing cycle. At the end of the preset time interval, the module's output switch closes until it is reset by removal of input power. Switch rating is 1/2 A resistive at 28 V dc.

CIRCLE NO. 274

MODULES & SUBASSEMBLIES

LED-fiberoptic display shows large characters


Combining 56-mil-dia optical fibers with LEDs, the 908 readout forms 0.85-in.-high by 0.43-in.-wide dot-pattern characters. The fiber-optics pipe over 99.5% of the illumination from the LED sources to a 7-segment character face providing 160-degree viewing. Contrast ratio is 50:1 (fiber-to-face) and each LED operates at 50 mA and 5 V dc. Current-limiting resistors are built in.

CIRCLE NO. 275

Seven-segment display plugs into 16-pin DIPs

Luminetics Corp., Box 1943, Pompano Beach, Fla. Phone: (305) 933-4551. Price: $2.90.

The series 90 MiNitron seven-segment incandescent display plugs into standard 16-pin DIP sockets. Dimensions of the display housing are 0.867-in.-high by 0.453-in.-wide by 0.024-in.-deep. Character height is 0.36 in. with a 5-degree slope. The display is rated at 5 V dc and 9 mA per segment. Filters are available in a variety of colors depending on application.

CIRCLE NO. 276

Differential amplifier slews at 30 V/μs


The Ampac MP215 differential inverting amplifier features open-loop gain of $2 \times 10^5$ and frequency response to 4 MHz, while slewing at 30 V/μs and settling to 0.01% in 1.5 μs. Its temperature coefficient is 20 μV/°C and input impedance is 1011 Ω. Input bias current is only 50 pA and noise is only 15 μV rms. Output is ±10 V at 20 mA and is short-circuit proof to ground.

CIRCLE NO. 277

Fast multiplexer is 0.01% accurate

Varadyne Systems, 1020 Turnpike St., Canton, Mass. Phone: (617) 828-6395. P&A: $198; stock.

Model MM-8 eight-channel analog multiplexer with a 1-μs settling time features an accuracy of 0.01%. All its addressing logic inputs are compatible with DTL/ TTL levels. Full-scale inputs can be either ± 5 or ± 10 V. Each channel can sequentially switch at a 500-kHz rate. The MM-8 is self contained in a plastic module measuring 0.8 cubic in.
Frequency converters mount on PC boards
Anadex Instruments, Inc., 7833 Haskell Ave., Van Nuys, Calif. Phone: (213) 873-6620. P&A: from $175; 4 wks.

The new series 1700-5096-XX of frequency-to-voltage converters on PC boards offer twelve input-frequency ranges that provide a 0 to 5-V dc output directly proportional to the frequency input. Ranges are from 100 Hz to 50 kHz.

CIRCLE NO. 279

Fast-settling op amp reaches 0.01% in 0.15 μs

A new FET fast-settling op amp, the FST-157, has an input impedance of 10¹² Ω, a dc gain of 300,000, a gain-bandwidth-product of 12 MHz and settles to 0.01% within 0.15 μs. Overload recovery time is 1 μs and frequency response is at a 6-dB/octave rate.

CIRCLE NO. 280

10-bit d/a converters cut settling to 200 ns
Teledyne Philbrick Nexus, Allied Dr. at Rte. 128, Dedham, Mass. Phone: (617) 329-1600. P&A: $70, $75; stock.

Models 4006 current and 4007 voltage 10-bit d/a converters with settling times of 200 ns and 25 μs, respectively, are contained in low-profile 2 by 2 by 0.4-in. packages. Both are DTL/TTL compatible and have DIP configurations.

CIRCLE NO. 281

Fast 75-MHz buffer gets to 0.1% in 40 ns
Intronics, Inc., 57 Chapel St., Newton, Mass. Phone: (617) 332-7350. P&A: $175; stock to 2 wks.

The model FA501 FET follower buffer with a risetime of 5 ns, an output of ±100 mA at ±5 V and a dc-to-75-MHz bandwidth settles in 40 ns to 0.1% of final value. It occupies 3.7 cubic in. of space and mounts on PC boards with relative ease.

CIRCLE NO. 282

More new drivers...
level adapter/level shifter

LMD-9 SHIFTER
LMD-10 ADAPTER
Now available from shelf

Typical Application and Connection Diagrams

Typical Specifications

<table>
<thead>
<tr>
<th>Voltage</th>
<th>30V</th>
</tr>
</thead>
<tbody>
<tr>
<td>Output</td>
<td>200mA</td>
</tr>
<tr>
<td>Total package power</td>
<td>450mw</td>
</tr>
<tr>
<td>Input</td>
<td>35V</td>
</tr>
<tr>
<td>Voltage drop</td>
<td>1.5V</td>
</tr>
<tr>
<td>Surge</td>
<td>0.75 Amp</td>
</tr>
</tbody>
</table>

Our LMD-9 is an interface quad driver and level shifter NPN circuit designed to solve your low-to-high voltage and current switching applications when the load is on the hot side.

When the load is on the ground side, use our LMD-10, a quadruple PNP adapter and driver, for your interface between low level integrated circuits and high voltage levels.

You can use the drivers with solid state stepping switches, to convert low level counter decoder outputs to levels compatible with your application. You can also use them as multiple position drivers, inverters to drive higher power transistors, or to form the electrical equivalent of NO or NC switch contacts.

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We're equipped to give you fast design and prototype service on any custom hybrid microelectronics package. Our engineers will come to you, if that's what you need.

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123 Webster Street, Dayton, Ohio 45401 phone (513) 222-6992
INFORMATION RETRIEVAL NUMBER 32
DIP packaging system accepts MSI LSI ICs


Dip-Mates are new dual-in-line packaging systems for MSI and LSI packages. Dip-Mates keep thermal problems to a minimum by air cooling the top and bottom of the plugged-in package. They consist of a thin voltage plate which is sandwiched to a base plate which functions as the ground plate. An insulated aluminum voltage plate contains a hard anodized finish.

CIRCLE NO. 283

Alumina substrates break-off in pieces

Ekon Electronics, 9760 Cozycroft Ave., Chatsworth, Calif. Phone: (213) 882-2063.

New scored alumina ceramic substrates with surface finishes of 15 μin. or less break cleanly into individual pieces after multiple circuit patterns have been deposited. The new substrates are available in sizes as small as 0.05 by 0.05 in. with 96% and 99.5% alumina compositions. Other compositions as well as colored ceramics for tracing and coding are also available.

CIRCLE NO. 284

Complete unsolder tool retails for $200


A portable unsoldering tool complete with a sprung component remover and a range of fittings for different component packages sells for under $200. The unit consists of an electrically heated pot of molten solder with a metal piston floating in it.

CIRCLE NO. 285

Resistance-wire kit spans 0.5 to 300 Ω/ft


Intended for making any value resistor, the Flextrim 101 low-cost kit contains both manganin and karma wire alloys in 7 resistance wires ranging from 0.5 to 300 Ω/ft (5000 Ω/ft optional). It also contains 11 epoxy glass bobbins.

CIRCLE NO. 286

High-temperature tape withstands 4000°F

Aremco Products, Inc., P. O. Box 145, Briarcliff Manor, N. Y. Phone: (914) 762-0685. P&A: $19.50 (1/2-in. 18-yard roll); stock.

Pyro-tape 546 is a high-temperature glass-plastic tape which can withstand continuous temperatures of 500°F and flash temperatures up to 4000°F. It is available 0.006-in. thick with an adhesive backing in 1/2 and 1-in. widths.

CIRCLE NO. 287

50-ppm resistive pastes have 1-MΩ square value


Resistivities up to 1 MΩ/square are available from the series 500 noble-metal resistive pastes with temperatures coefficients of 50 ppm/°C. These pastes are recommended for firing conditions of 875°C for eight minutes at peak temperature.

CIRCLE NO. 288

Evaluation samples

Pushbutton switches

Free samples of lighted pushbutton switches are available in personalized colors to match any equipment. The color-dyed switches, designated the 1820 series, may be ordered in any color or color combination for the switch parts of the button, nylon bezel and housing. Metal bezels come in chrome, black oxide or brass-plated steel. The 1820 line is rated for 2 to 9-A service at 125 or 250 V ac. Options include spst or spdt models. All switches have built-in diffusing screens to produce a glow light without an objectionable hot spot. Molex, Inc.

CIRCLE NO. 289

Harness ring

A new open-connector harness ring permits a harness to be added to existing wire bundles without installing additional cable ties to the initial bundles and without cutting off the original ties. The CROS ring is installed with a cable tie on the auxiliary harness. A slot in its top permits it to be fastened to any previously installed miniature, intermediate or standard cross-section cable tie. Free samples are available. Panduit Corp.

CIRCLE NO. 290
Tolerance comparator

The True Position Tolerance Comparator is a 4-1/2 by 11-in. plastic sliding-scale device that provides instant analysis of true position tolerancing and dimensioning. The Inspector’s Aid side of the comparator shows coordinate measurements with respect to circular tolerance zones on the drawings. The Manufacturer’s Aid side shows how far from true position the tool settings may deviate. The comparator comes with an instruction manual and costs $7.50. TAD Products Corp.

Silicon solar cells

A four-page application report thoroughly discusses the physical and electrical properties of silicon solar cells. Several curves illustrate the cells’ different characteristics. AEG-Telefunken Corp.

Beam-lead sensors

How a designer can achieve maximum design freedom and product reliability by employing a beam-lead sensor array instead of a large number of individual discrete phototransistors is discussed in an application report. The report explains beam-lead sensor array technology and the advantages of such an array compared to one comprising individual phototransistors mounted and wired separately. Texas Instruments Inc.

Trapatt diodes

A three-piece set of application literature on an avalanche trapatt diode is available. General Electric.

Mini / Bus by Rogers

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A small, voltage-distributing busbar for PC card application, each Mini/Bus gives you built-in capacitance... noise-cutting capacitance that means more reliable, compact circuit packaging at a fraction of multilayer prices. Write for data.

Rogers Corporation / Rogers, Conn. 06263
new literature

Computer handbook
A 300-page handbook describing the hardware and software features and uses of PDP-12 laboratory computer systems is available. The handbook is designed as a basic guide for users of PDP-12 systems and as an aid to individuals who are considering the acquisition of a laboratory computer. It contains sections on biomedical, chemical, educational, industrial and clinical laboratory computer systems. Digital Equipment Corp.

CIRCLE NO. 296

Potentiometers
A four-page brochure describes conductive plastic potentiometer elements and gives information on their electrical parameters. Waters Manufacturing, Inc.

CIRCLE NO. 297

Video switchers
Distinctive features of new production video switchers are listed in detail in a six-page two-color technical data sheet. Cohu Electronics, Inc.

CIRCLE NO. 298

IC rework
Effective IC repair, rework and modification of advanced electronic assemblies is the subject of an eight-page brochure. Paei Inc.

CIRCLE NO. 299

Recorders
A four-page illustrated brochure describes portable and rack-mounted general-purpose recorders. Gould, Inc.

CIRCLE NO. 320

Digital recorders
A four-page brochure describes a cassette magnetic-tape operating system for mini-computers. Dicom Industries

CIRCLE NO. 321

Imaging devices
A new quick-selector guide provides rapid and comprehensive reference to a full line of imaging devices. Products are discussed from a performance vs application point of view. Westinghouse.

CIRCLE NO. 322

Sound system components
A six-page catalog gives detailed specifications and descriptions of a line of commercial sound components and special-purpose sound system products. Bell P/A Products Corp.

CIRCLE NO. 323

Tape readers
A four-page brochure describes photo-electronic punched-tape readers capable of reading speeds up to 600 characters/s. Decitek Div. of Jamesbury Corp.

CIRCLE NO. 324

Hobbyist catalog
An 80-page hobbyist catalog includes watches, geiger counters, logic cards, electronic kits, cameras, lenses and many more items of interest. B. & F. Enterprises

CIRCLE NO. 325

MOS/LSI ICs
MOS LSI shift registers, logic cards, memories and custom circuits are described in a brochure. Philco-Ford.

CIRCLE NO. 326

Temperature indicators
A new comprehensive four-page catalog describes a line of compact digital temperature indicators. Thermo Electric.

CIRCLE NO. 327

Analog multipliers
Two foldout data sheets describe a pair of internally trimmed and wideband transconductance analog multipliers. Analog Devices Inc.

CIRCLE NO. 328

Rocker switches
A four-page data brochure describes rocker and paddle-handle switches. It contains photographs, schematics, specifications, ordering information and prices. C & K Components, Inc.

CIRCLE NO. 329

Capacitors
Two catalogs cover a full line of capacitors. The line includes dipped silver mica, aluminum electrolytic, MIL-approved ceramic and polyester film capacitors in all sizes. Miconics Industries, Inc.

CIRCLE NO. 330

How long must you keep important papers?
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CIRCLE NO. 325

INFORMATION RETRIEVAL NUMBER 35

Electronic Design 3, February 4, 1971
GC Electronics catalog

The 1971 edition of the GC Industrial Electronic components catalog containing nearly 1700 electronic items is available. The 84-page catalog lists connectors, adapters, alignment tools, clips, plugs and jacks, binding posts, cable clamps and cable ties. Chemicals, nylon and metal hardware, PC materials and connectors, spacers, switches, test probes, wire strippers, grommets and lacing cords are also included. GC Electronics Division of Hydrometals, Inc.

CIRCLE NO. 331

Connectors

A new connector that meets MIL-E-5400K requirements is described in a four-page bulletin. Microdot, Inc.

CIRCLE NO. 332

Cable ties and clamps

A brochure describes a full line of one-piece cable ties, clamps and marker ties. Panduit Corp.

CIRCLE NO. 333

Wire and cables

Three wire and cable lines are described in a new catalog. The three lines feature flexibility and durability. Calmont Engineering and Electronics.

CIRCLE NO. 334

Magnetic alloy

A soft magnetic alloy for operation at high flux densities in motor, generator and transformer laminations, torque devices, pole pieces and solenoid cores is described in a new brochure. Westinghouse Electric Corp.

CIRCLE NO. 335

Price reductions averaging 10% on SSL series solid-state lamps have been announced by General Electric.

CIRCLE NO. 338

Potter Instrument Co., Inc., Plainview, N. Y. announced a price reduction on its SC-1035 single-capstan tape transport from a previous price of $3800 each (100-unit quantities) to a new price of $2850 each (100-unit quantities).

CIRCLE NO. 339

Price reductions on three LEDs ranging up to 45% were announced by Litronix, Inc. The three devices are the Red-Lit 2, 4 and 7 LEDs. Each of the three LEDs now costs as low as 99c.

CIRCLE NO. 340

RCA Electronic Components has announced price increases ranging from 3 to 7% on five sizes of color TV CRTs.

CIRCLE NO. 341

Design Data from Manufacturers

Advertisements of booklets, brochures, catalogs and data sheets. To order use Reader-Service Card.

(Advertisement)

Highest Resolution in New Spectrum Analyzers

For the on-line processing of underwater acoustic signals, radar, sonar and other data where the maximum frequency resolution is always sought, new UA-9A series 1,000-line and 2,000-line Low Frequency Real-Time Spectrum Analyzers have been designed by Federal Scientific, originators of the Ubiquitous® analyzers. High-speed versions produce complete 1,000-line spectra in 40 msec, and 2,000-line spectra in 160 msec. And now any UA-6A series 500-line Ubiquitous analyzers in the field can be expanded to the factory to 1,000-line or 2,000-line performance.

CIRCLE NO. 171

Federal Scientific Corporation
a subsidiary of Elgin National Industries, Inc.
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The Transferred-Electron Amplifier (TEA) is the first wideband microwave amplifier with potential to replace bulky vacuum devices with compact, lightweight, highly-reliable units for communications, ECM, and airborne phased arrays.

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