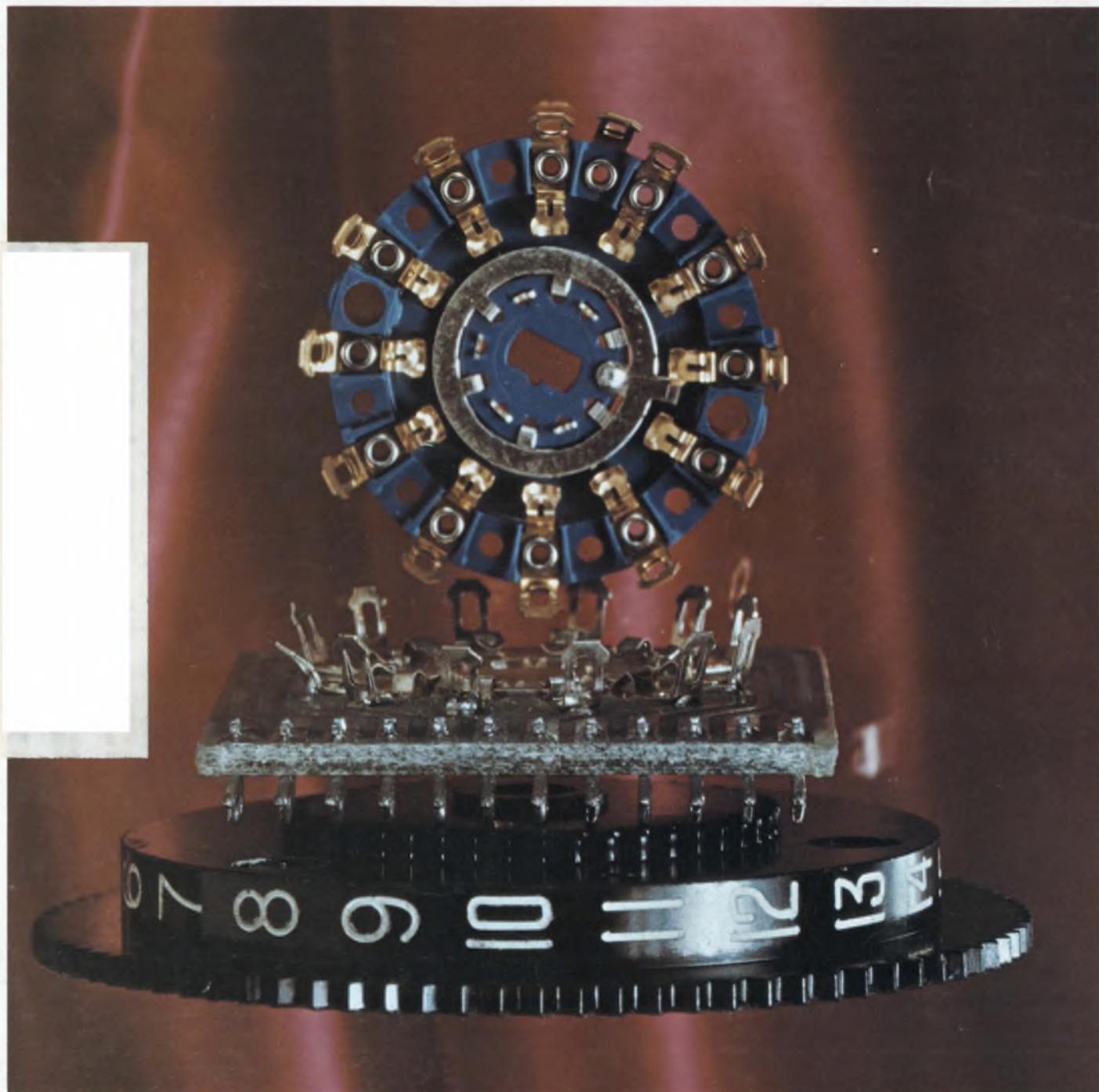


Wound up over rotary switches?

You can choose among twist, thumb, push and lever styles. Which should you pick? Twist types remain the workhorses,

thumbwheels dominate digital input uses and some types are designed for circuit-board use. But catalogs don't reveal enough. To help unwind, turn to page 56.





Ultra reliability in
ultra-miniature
transformers...

...and they're off the shelf!

PICO Electronics' patented construction combines mass production with the utmost in reliability to produce the smallest possible transformer.

These little giants handle up to 600 milliwatts at 1 KHz and 1.5 watts at 10KHz. They cover the frequency range of 300Hz to 250KHz. Primary or Secondary impedances of 3.2 ohms to 250K ohms. Pulse applications .05 μ s to

100 μ s. Construction can be TO-5 plug-in, insulated flexible leads or flat pack ribbon style.

Size variations from $\frac{1}{4}$ " dia. x $\frac{1}{4}$ " ht. to $\frac{1}{32}$ " dia. x $\frac{1}{2}$ " ht. Special designs with $\frac{1}{4}$ " dia. and .180" ht.

Pico Electronics is a Q.P.L. source. All transformers are hermetically sealed in a metal case and in strict compliance with MILT-27 specifications. Every Pico device has the added reliability feature of being

guaranteed for 25 cycles of thermal shock (MIL-STD-202D, method 107C).

Industrial and Commercial designs are also available for maximum economy.

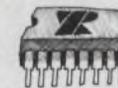
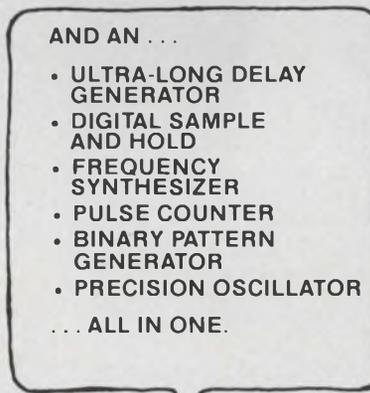
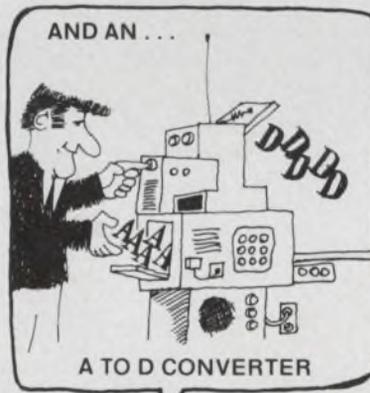
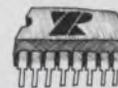
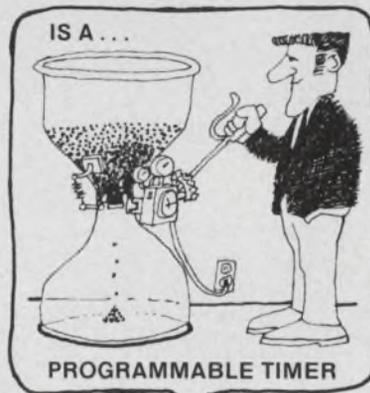
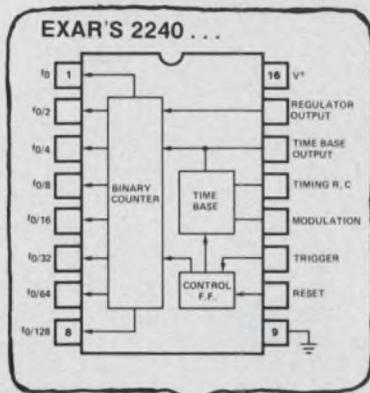
Over 600 standard transformers to choose from... Delivery - stock to one week... specials take a little longer - 9 days.

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THE FIRST OF THE BIG COUNT TIMERS



Exar's new XR-2240 counter/programmable timer solves so many tough problems that designers will unanimously agree that it's really the universal timer.

With its unique combination of analog and digital timing methods, you can now replace inadequate and complex assemblages of monolithic and electromechanical timers with the much simpler XR-2240. As a bonus, you get greater flexibility, precision operation, and a reduction in components and costs for most applications.

Because of built-in programmability, you can also use the XR-2240 for frequency synthesis, electronic music synthesis, digital sample and hold, A to D conversion, binary counting and pattern generation, and more.

With a single XR-2240 you can now generate

precision time delays programmable from 1RC to 255RC, a range of microseconds to 5 days. By cascading only two XR-2240 timers, you can extend the maximum delay by a factor of 2^N , where $N = 16$ bits, resulting in a total delay of 3 years!

The XR-2240 operates over a 4V to 15V supply range with an accuracy of 0.5% and an 80 ppm/°C temperature stability. It's available in either a 16-pin ceramic or plastic dual-in-line package for military or commercial applications. Prices start at \$4.50 in 100 piece quantities.

For the more conventional timing applications, look to our other timers: the XR-320 timing circuit and the XR-2556 dual timers. Call or write Exar, the timer leader, for complete information.

EXAR SPEAKS YOUR LANGUAGE

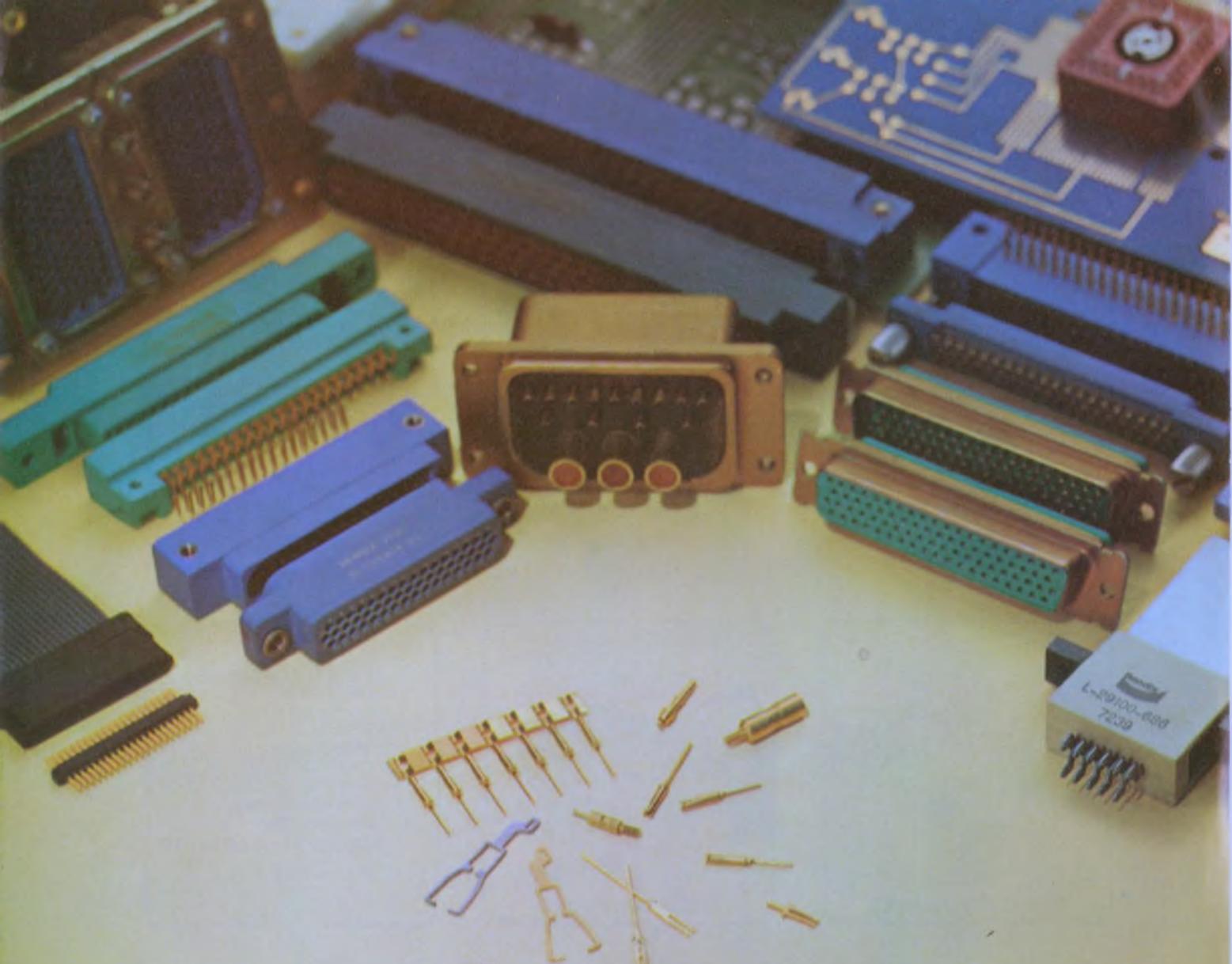
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INFORMATION RETRIEVAL NUMBER 2



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You can choose from a wide range of contacts as well. Crimp-type rear removable. Solderless wrap. Solder termination. Floating. Straight or right angle.

Choose from a wide range of connector styles and contact variations.

All this means if you're in computer main frame, communications equipment, avionics or automotive electronics, Bendix probably has just the connector you need.



Fact is, we could write a book about our square and rectangular connectors so we've done precisely that. Why not send for it? It gives you over 70 pages of ways to go square and rectangular, too. Write: The Bendix Corporation, Electrical Components Division, Sidney, New York 13838.



NEWS

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- 32 **An ultra-low-frequency electromagnetic direction-finding system** for miners passes tests at 1500-ft depth.
- 36 **Hand-held gas detector** promised, with 1-part-per-billion sensitivity.
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- 70 **Design of oscillators** from transistor data sheets. By using simple models you can easily compute the component values for sinusoidal oscillators.
- 76 **Convert two-port circuit parameters** from any form to any other. A short Fortran program handles all popular parameters in polar or rectangular form.
- 82 **Analyze TV noise performance** by correlating video signal-to-noise ratio with rf noise figure. Then use a spectrum analyzer to perform the measurements.
- 90 **ECL 10,000 interconnects economically.** Controlled edge speeds and low-voltage swings allow use of conventional boards for this popular logic.
- 98 **Look as asynchronous sequential logic** for best speed-power performance. Then add a self-clocking feature to obviate design constraints.
- 106 **Is a crowbar alone enough?** To reduce the risk of power-supply damage or even fires, an overcurrent interrupt element may also be needed.
- 114 **Servicing a high-technology product** means that engineering must provide a design capability that minimizes later problems with the customer.
- 118 **Ideas for Design:** Counter has symmetrical output though the input signal is asymmetrical . . . Simple tally system ranks events in the order of occurrence . . . Inexpensive AM modulator replaces clipping type and gives less distortion . . . Comparator detects volts in narrow window and accurately nulls to reference level . . . Counter resets itself reliably with one additional flip-flop.
- 128 **International Technology**

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Cover: Photo by Fred Meyer, courtesy of OAK Industries Inc., Switch Div.

“Think Hz for a

“A different point of view . . . that’s often the key to a problem solution. And that’s just what HP analyzers give you. A look at your problems in the frequency domain instead of the time domain.”

“HP high-frequency analyzers have a solid reputation for providing RF and microwave designers with accurate frequency-domain measurements. But what about analyzers to help engineers with low-frequency problems?”

“I’d like to tell you about HP’s broad line of low-frequency analyzers that can solve some of your tough measurement problems. Problems such as making low-level measurements in

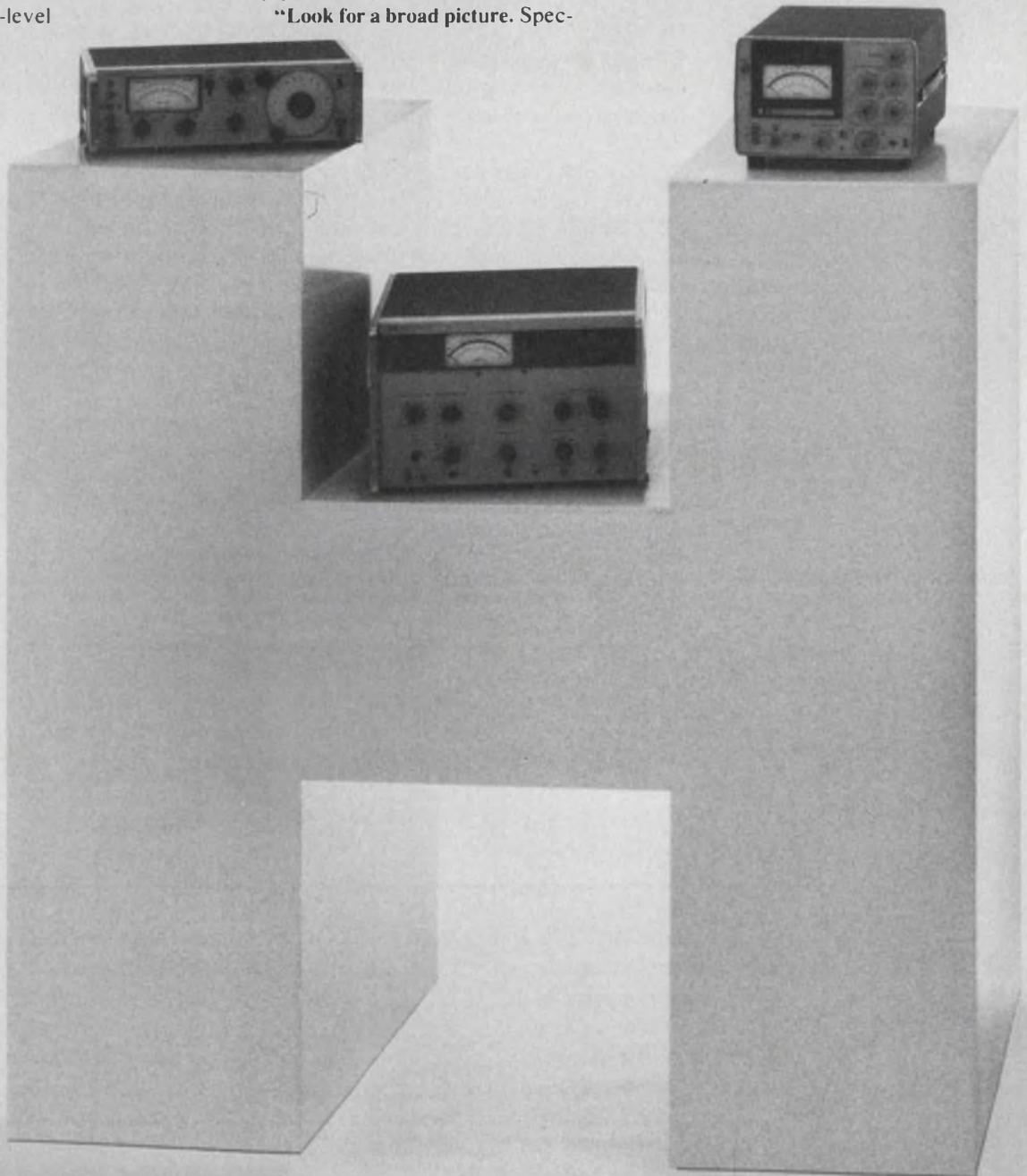
the presence of noise, analyzing frequency components of a signal, determining the transfer function of a network, or making distortion measurements on analog circuits.”

“Take the narrow view. HP’s 3581A wave analyzer lets you home-in on a specific frequency for amplitude measurements. Look at a signal through a 3 Hz bandwidth. Monitor low-level signals to 30 nV over a 15 Hz to 50 kHz range. This ability to measure low-level signals in the presence of noise simplifies transducer signal monitoring. Helps you isolate ground-loop problems.”

“Look for a broad picture. Spec-

trum analyzers display your waveform’s Fourier components on a CRT. That’s an easy way to see your signal and its component relationships. It’s also an ideal way to characterize mechanical devices that have unique spectral ‘signatures.’”

“HP’s 3580A spectrum analyzer gives you a 5 Hz to 50 kHz range and the sensitivity to operate directly from the lowest level transducer outputs. Its digital memory lets you store traces for accurate comparison of two signals. And a narrow 1 Hz bandwidth allows you to resolve 60 Hz sidebands.”



different view™

“There’s also an in-depth view provided by HP’s network analyzers. These dual-channel analyzers provide amplitude and phase measurements as a function of frequency. This simplifies transfer-function measurements, speeds the location of poles on a Bode plot, and accurately determines resonant frequencies and phase margins in circuit design.”

“Our broadband 3575A gain-phase meter gives you a range of 1 Hz to 13 MHz and lets you select your own signal source to do low-cost network analysis.”

“For high-accuracy network analysis, plus optional group delay and offset determinations to 13 MHz, HP’s high-performance 3040 analyzer is the answer. By adding the computational power of a calculator

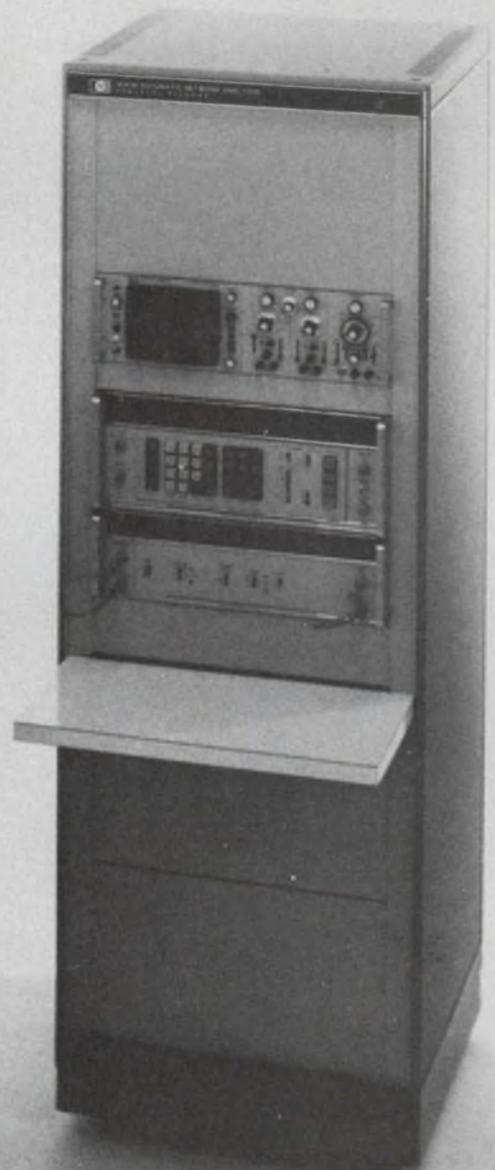
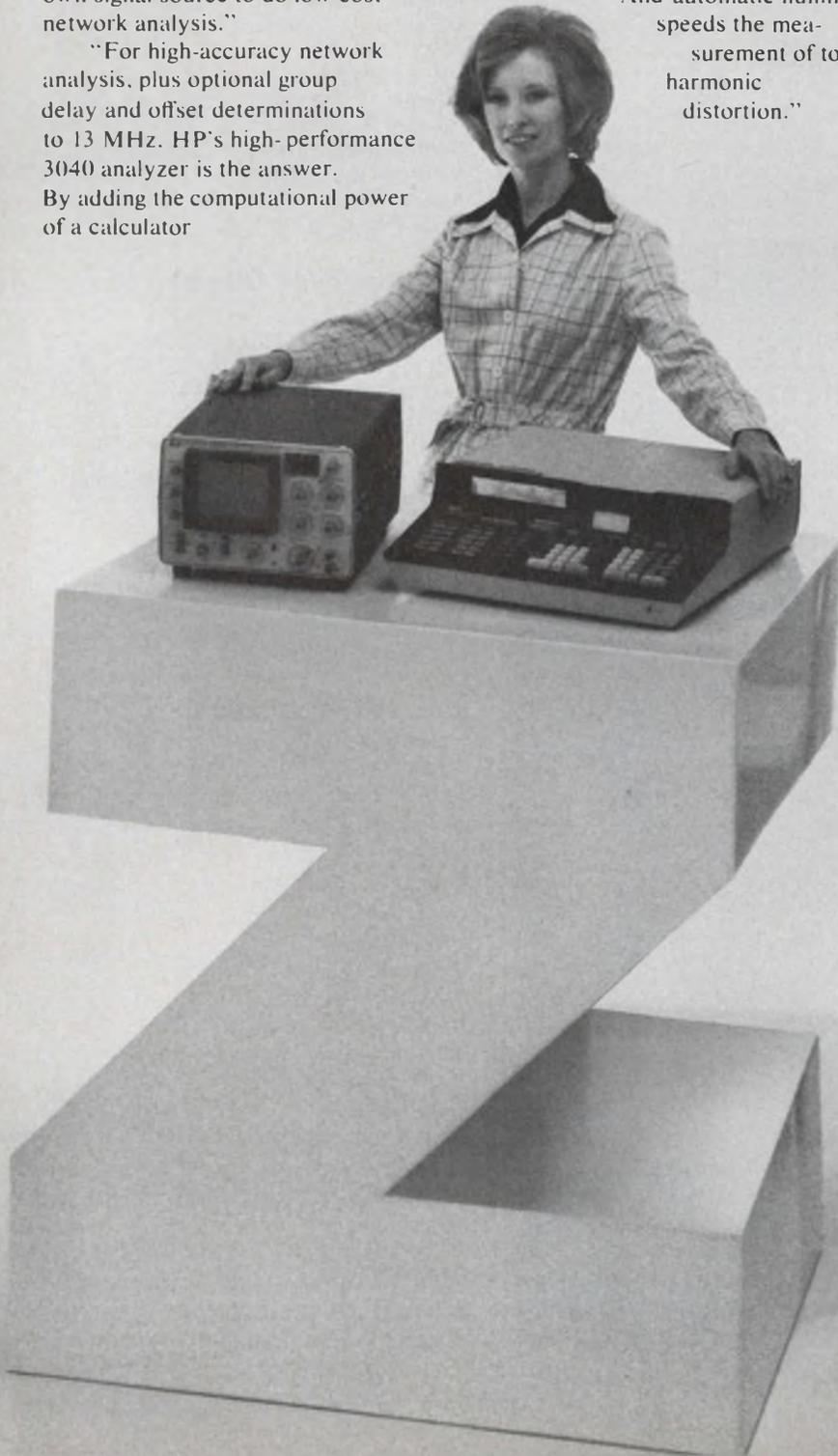
to either the 3575 or 3040 analyzers you can view the results of 108 tests in less than 20 seconds... and with greater precision.”

“And there’s a reciprocal view. HP’s distortion analyzers give you the reciprocal function of a wave analyzer. By rejecting the fundamental and monitoring the remaining harmonic and spurious components, distortion is measured and displayed on a meter. Our 331A to 334A series provides a wide range of 5 Hz to 65 MHz.

And automatic nulling speeds the measurement of total harmonic distortion.”

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across the desk

Engineers! Join IEEE for a new status!

If you're an IEEE member, there's a constitutional amendment in your future. In fact there are two constitutional amendments that will be up for approval on the 1974 ballot. Both have come up the hard way—by petition from the Professional Activities Committees (PAC) throughout the U.S.

In 1972 the constitution was amended to permit the IEEE to take action on behalf of the professional status of its members. To implement this policy, the headquarters encourages each section (of which there are about 260) to form a PAC to handle the professional activities of the section. The PAC studies engineers' professional problems and makes recommendations or takes action to ameliorate the problems.

The Long Island PAC has come up with a petition to pay the IEEE president a salary, to give him a three-year, full-time term of office—instead of one year—and to obligate him to disaffiliate completely from the electrical or electronics industry during his term of office. His salary would be competitive with executives of like responsibility in private industry. We feel that this is a way to make him responsive to the membership and would end any conflict of interest between the presidential office and an outside job. Our PAC chapter circulated this petition throughout the country. We received far more than the required 450 signatures to put it on the ballot.

A California PAC performed the same process with respect to another amendment, which would

lower the number of signatures needed on a petition to put a second candidate's name on the ballot for president. The constitution formerly required signatures of 2% of the voting membership (about 2700), but this amendment would lower it to 0.33% (about 450).

Notice how nicely our two amendments fit together. Considering that PACs have been in existence less than two years, this is quite an accomplishment. You ain't seen nothing yet!

Position papers will accompany the ballot and the amendments and will tell why members should vote yes—if it's not immediately obvious. Long Island's PAC urges all voting IEEE members to vote yes on both proposed amendments. Democratize the IEEE with your ballot. Also watch for further action from PACs around the country.

So you see, the IEEE is on its way to becoming responsive to the needs of the working engineer. PAC is the avenue. The message for non-IEEE members is: Join IEEE and participate in the changeover. In so doing, you'll strengthen IEEE and your professional stature.

Robert Bruce

Chairman, L.I. Section PAC
15 Johnstone Rd.
Great Neck, N.Y. 11021

Clearly not clairvoyant

In these times of product and project planning, it is well to note that last month's meeting of the Southern California Clairvoyant Society was cancelled due to unforeseen circumstances.

Thin-Trim[®] capacitors



Tucked in the corner of this Pulsar Watch is a miniature capacitor which is used to trim the crystal. This Thin-Trim capacitor is one of our 9410 series, has an adjustment range of 7 to 45 pf., and is .200" x .200" x .050" thick. The Thin-Trim concept provides a variable device to replace fixed tuning techniques and cut-and-try methods of adjustment. Thin-Trim capacitors are available in a variety of lead configurations making them very easy to mount.

A smaller version of the 9410 is the 9402 series with a maximum capacitance value of 25 pf. These are perfect for applications in sub-miniature circuits such as ladies electronic wrist watches and phased array MIC's.

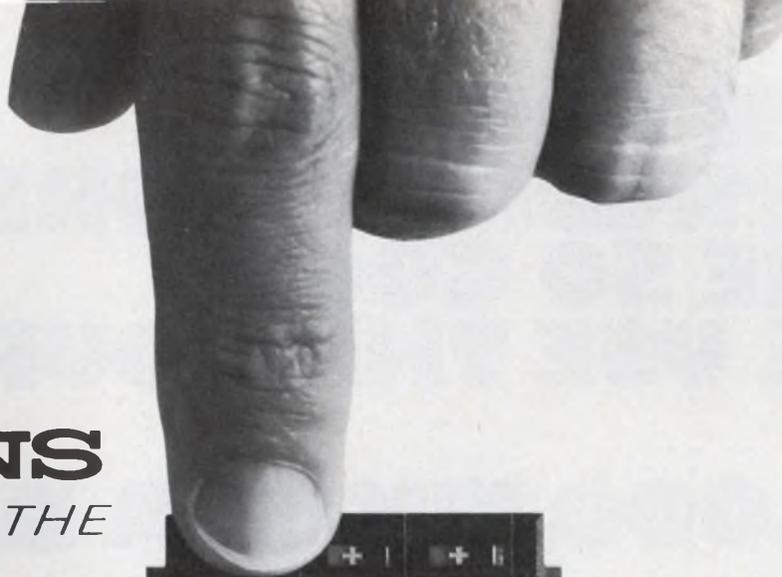
Johanson Manufacturing Corporation,
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07005. Phone (201) 334-2676, TWX 710-
987-8367.

Electronic Design welcomes the opinions of its readers on the issues raised in the magazine's editorial columns. Address letters to Managing Editor, Electronic Design, 50 Essex St. Rochelle Park, N.J. 07662. Try to keep letters under 200 words. Letters must be signed. Names will be withheld on request.

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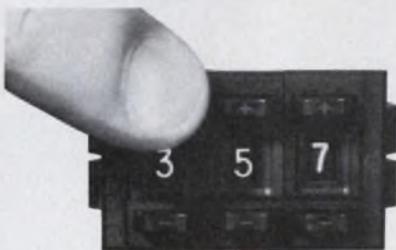


BOURNS INTRODUCES THE FIRST INTEGRAL PUSHBUTTON POTENTIOMETER



If you're designing panels with precision data entry or set-point controls, consider the BOURNS Model 3680 KNOBPOT® Digital Potentiometer . . . another innovative idea from Bourns. The 3680 integrates a precision incremental decade potentiometer with an easy-to-read digital display, AND a speedy pushbutton control action. It is handsome, extremely accurate, and a "snap" to install. Everything is INSIDE the Model 3680 . . . no resistors or mini-PC boards are required . . . nothing clutters the back of the unit to steal precious space.

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Simple, fast, precise. Push the PLUS button to increase; the MINUS to decrease. Rated life is 100,000 operations per decade.

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You get what you set with the 3680 . . . every time. The unique Bourns design integrates precision laser-trimmed cermet resistor technology with a positive pushbutton detent action. The result is resolution of output of 1 part in 1000 discrete steps, and dependable repeatability of $\pm 0.1\%$.

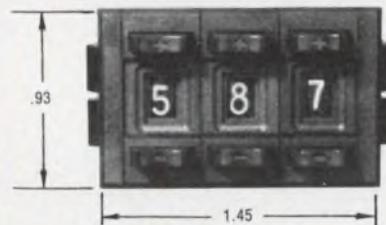
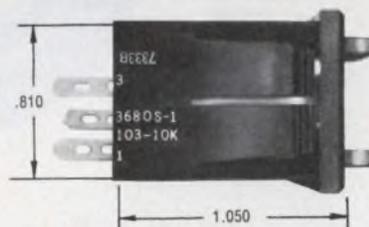
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Snap-in mounting cuts installation time, eliminates mounting hardware. Integral bezel covers irregular panel cut outs and minor edge blemishes. Terminals match the AMP Series 110 receptacle . . . or can be soldered in the standard fashion.

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- stable built-in cermet resistance elements
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INFORMATION RETRIEVAL NUMBER 7

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20,000 TIMES A DAY.**



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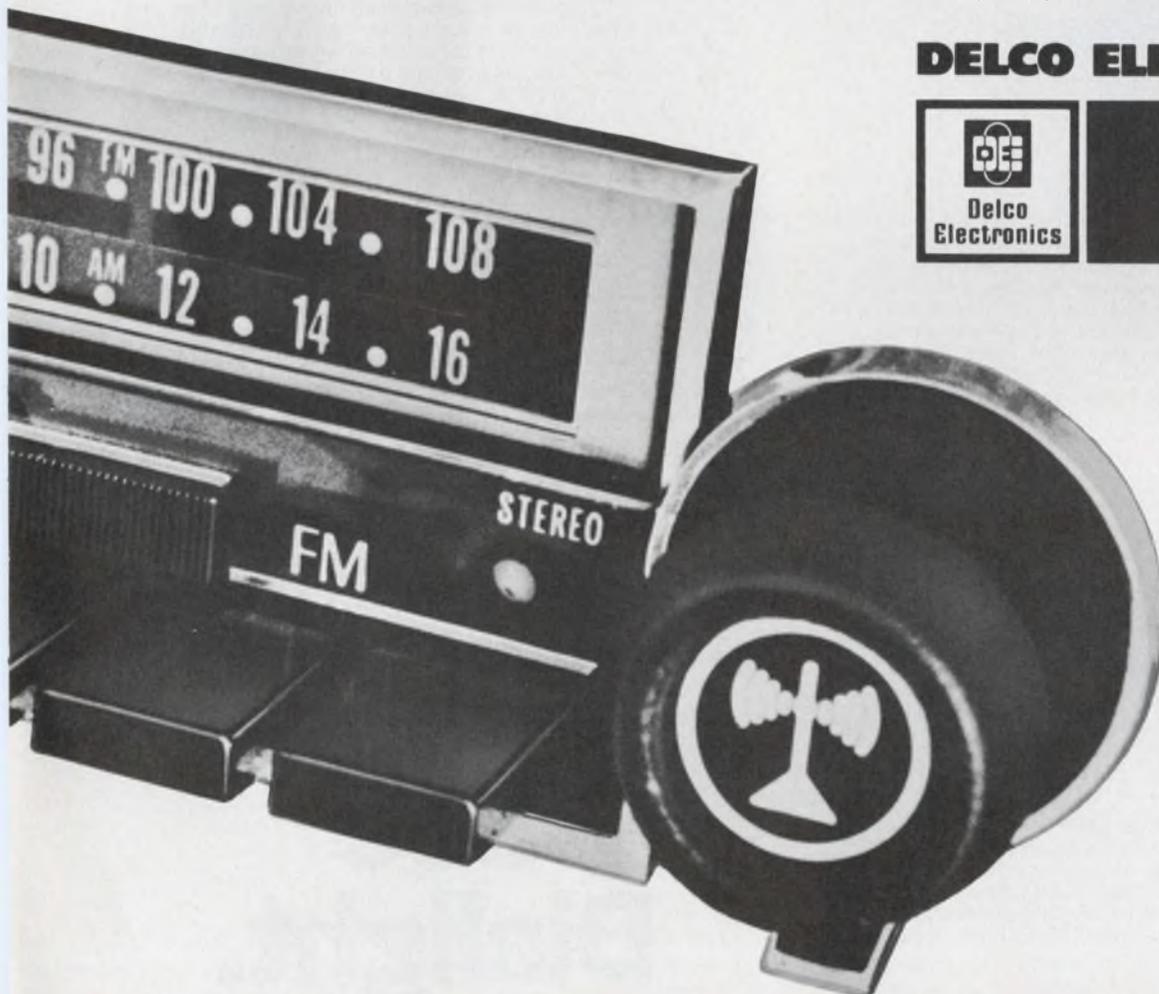
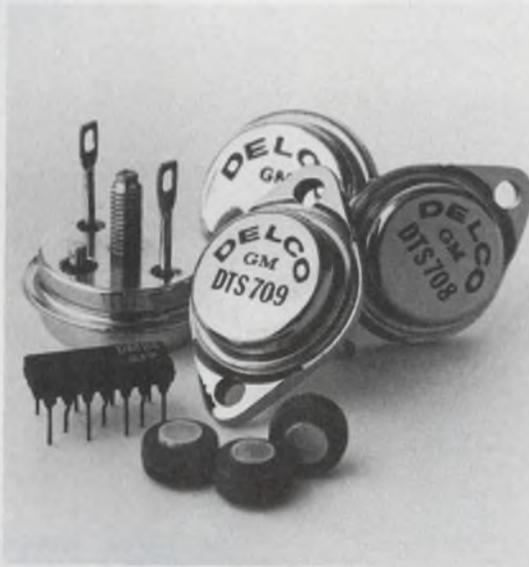
Our semiconductor devices are used

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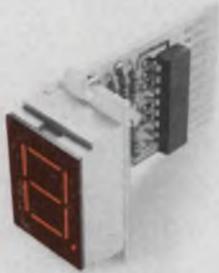


DELCO ELECTRONICS



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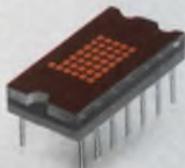
(Need: The widest choice for your every application.)



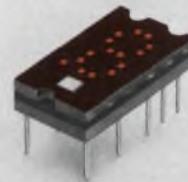
730 SERIES Your choice . . . a red or green LED readout with large 0.625" characters . . . low power, operates with standard IC power supply levels. Comes in plus-minus module. Display uses standard or high brightness LEDs for maximum light output arranged in a seven-segment format. Available with or without on-board decoder/driver. Unique lens design generates bright, highly legible characters.



739 SERIES Save design time and installation costs . . . this LED display assembly is attractively designed in a convenient package with bezel and is ready for instant panel mounting. Available in groups of one or more characters, with or without decoder/driver . . . characters are 0.625" and come with either green or red LEDs in seven-segment format. Readout offers lowest cost per character for comparable size.



745-0005 Dot matrix LED readout display produces a bright 0.300" high character or symbol display. Has wide angle visibility and is compatible with USASCII and EBCDIC codes. Low power requirements. Mounts into standard 14 pin DIP socket. The display is also available in bezel assemblies with or without code generator.



745-0007 LED hexadecimal display with on-board logic operates from 5 to 6 volt supply, low power consumption. Integral TTL MSI chip provides latch, decoder and drive functions. 0.270" character display has wide angle visibility and mounts into standard 14-pin DIP socket.



755 SERIES High brightness planar gas discharge displays in a 0.550" character. Orange color gives high contrast ratio and allows readability to 40 feet even in high ambient lighting. Designed for interfacing with MOS/LSI, displays have an expected life of 100,000 hours or more.

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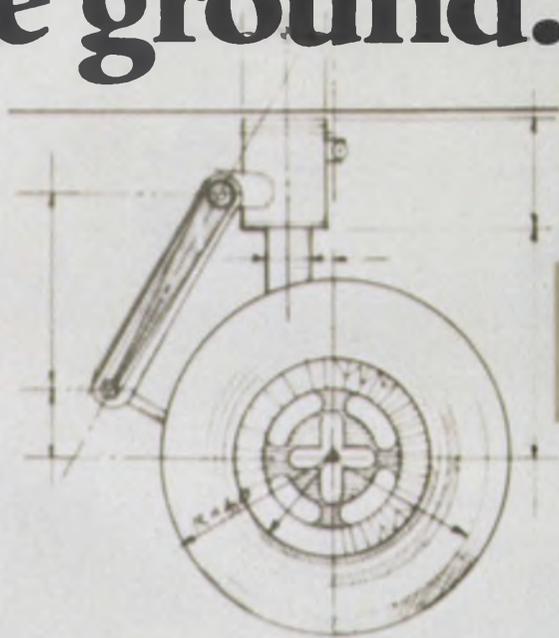
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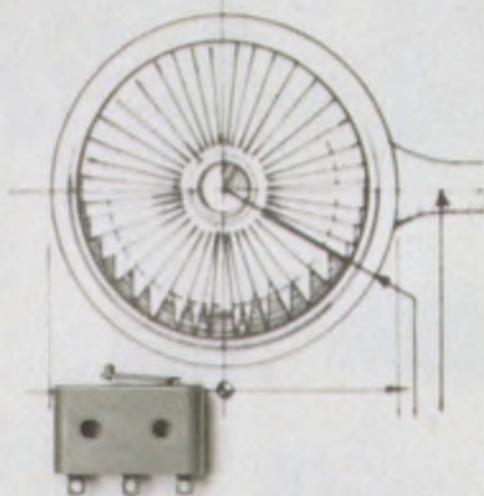
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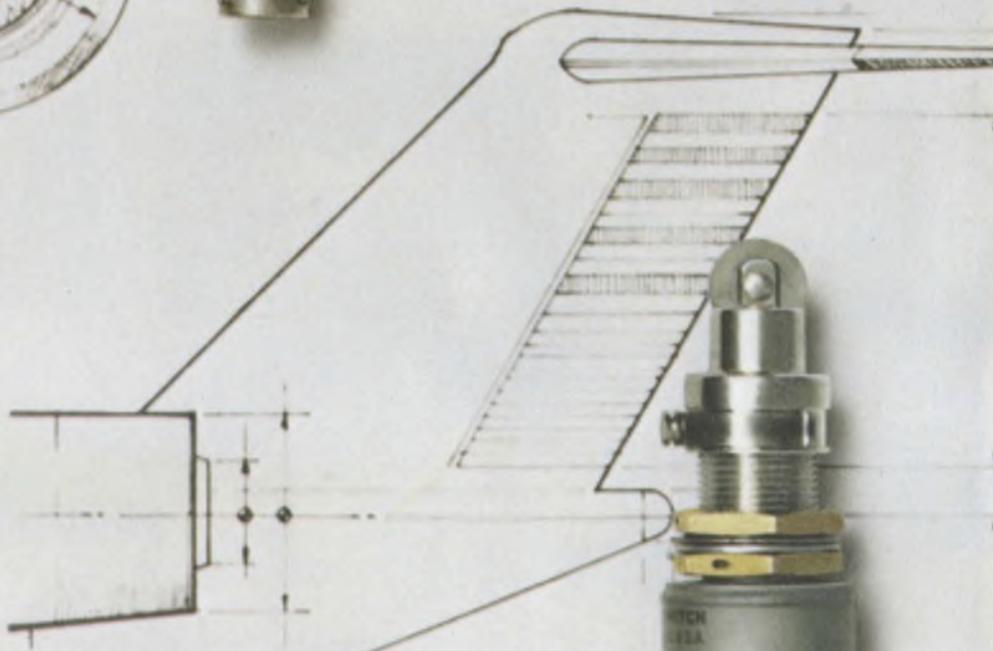
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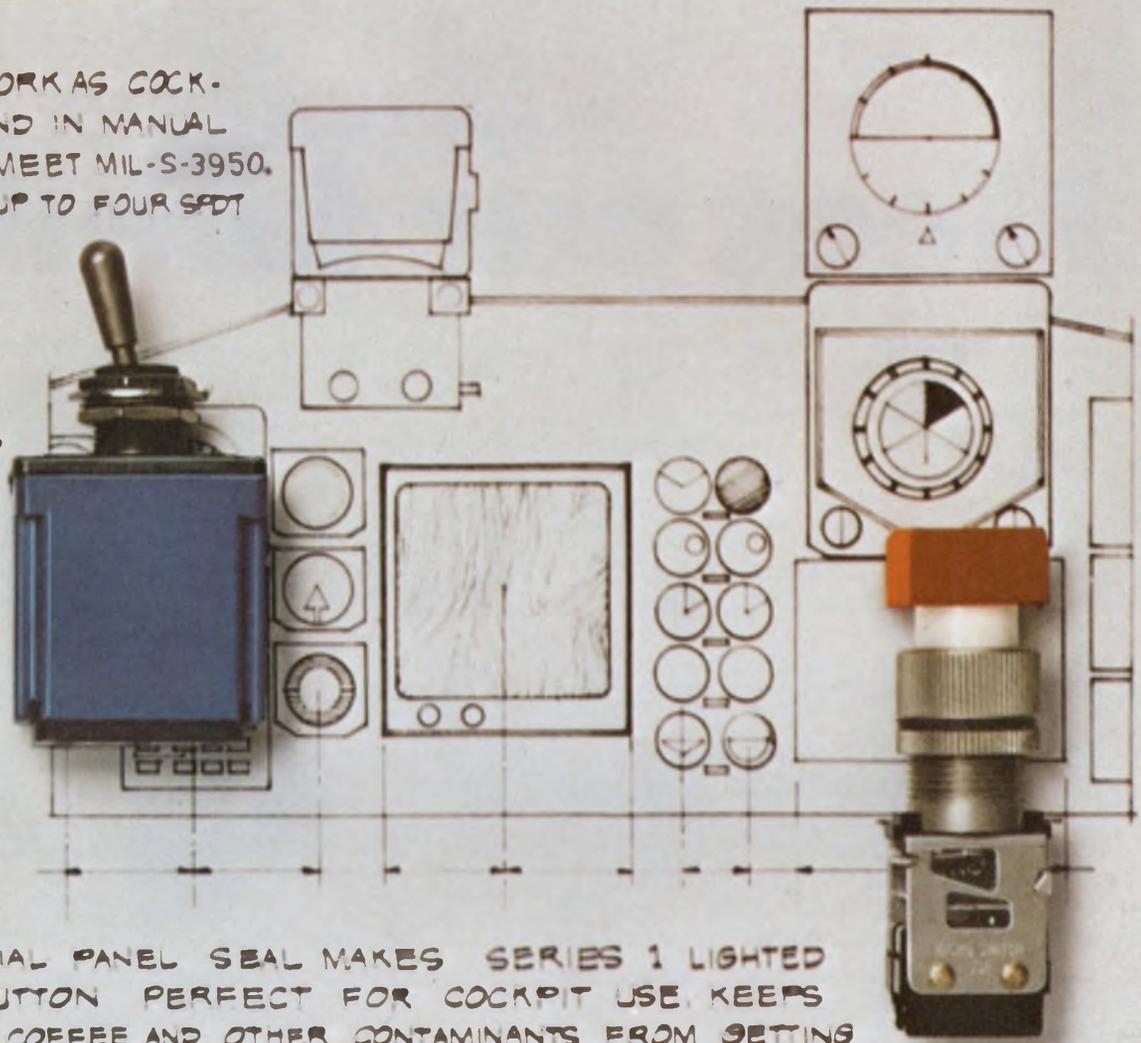


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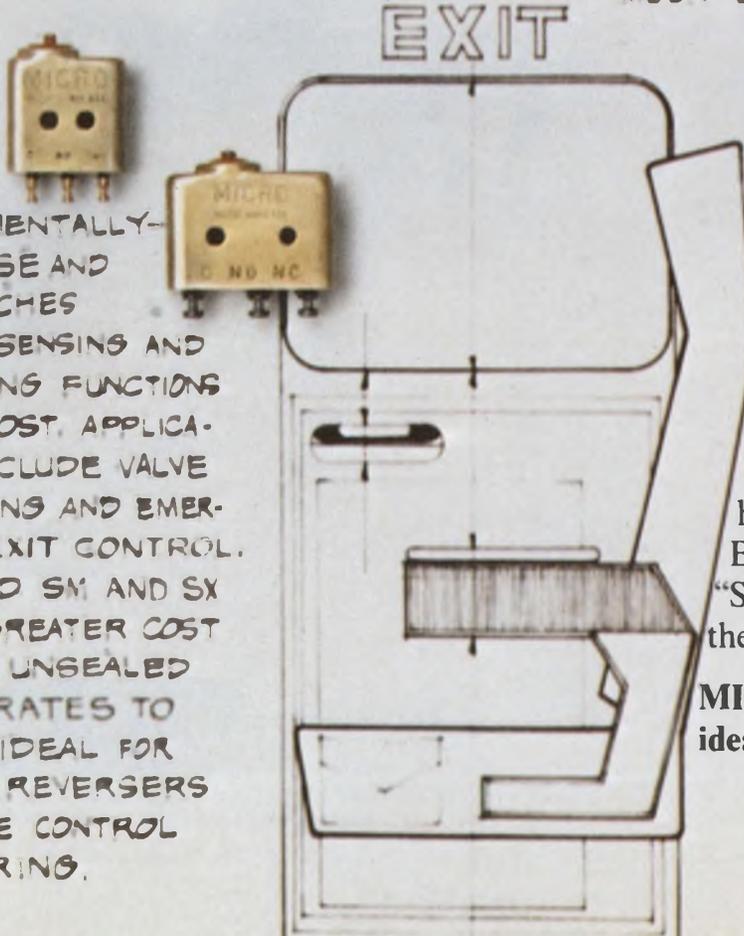
ENVIRONMENTALLY-SEALED EN'S CAN SENSE POSITION OF DOORS & CONTROL SURFACES, INDICATE STATUS OR INITIATE ACTUATOR CONTROL. OFFER WIDE ELECTRICAL RANGE, CHOICE OF TERMINATIONS & OPERATING CHARACTERISTICS. COMPLEMENTARY PRODUCTS OFFER HERMETIC SEALING AND TEMPERATURE RANGES TO $+600^{\circ}$ F.

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 INFORMATION RETRIEVAL NUMBER 11

Microwave Mixers

Meet our family.

Since one kind of mixer can't meet all kinds of needs, Watkins-Johnson has developed an entire family of double-balanced microwave mixers featuring:

High Isolation and Low Conversion Loss. The WJ-M14, M14A and M20 series provides >30 dB of isolation and <6 dB of conversion loss with $+7$ dBm of local oscillator drive across C, X, and Ku-band.

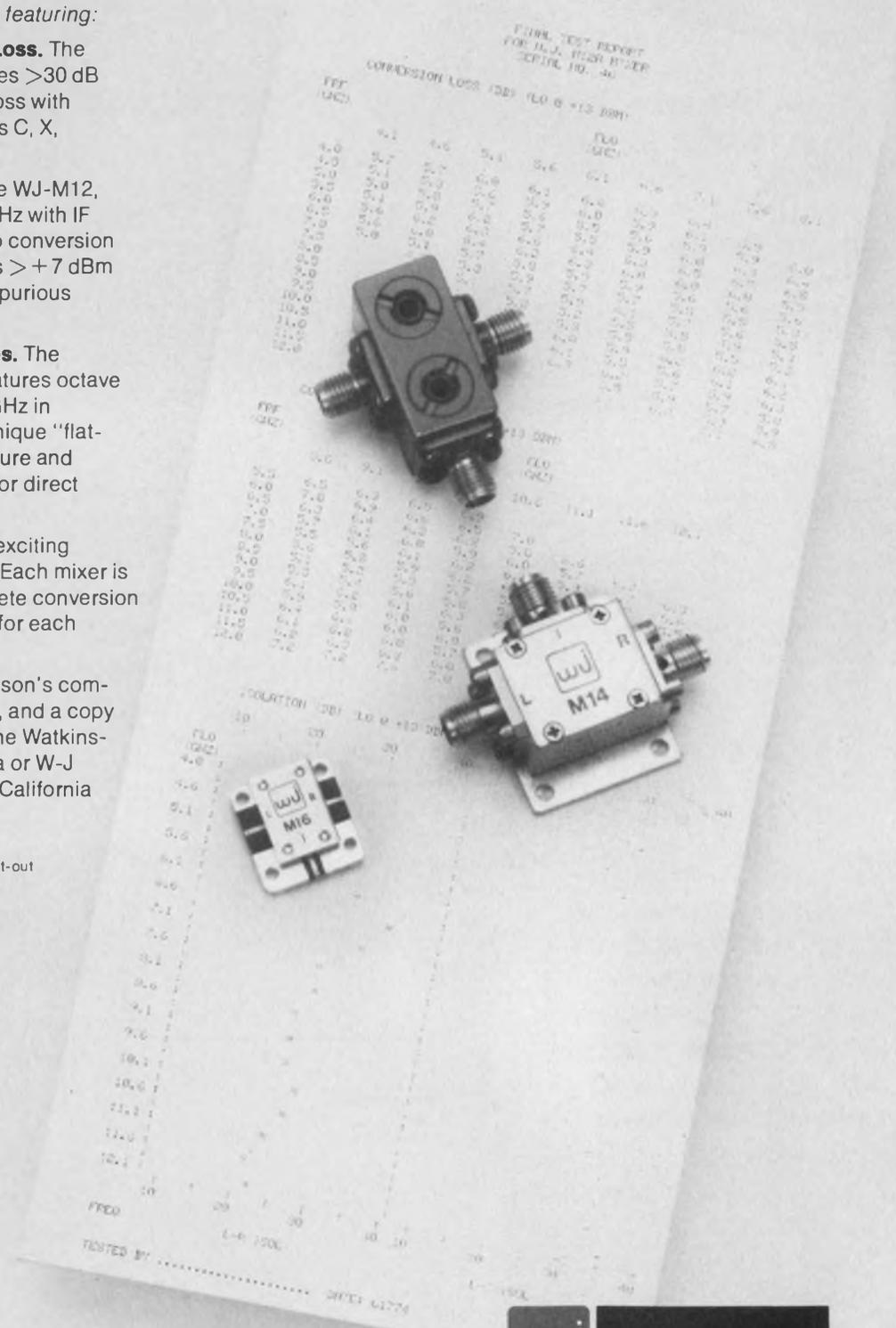
Very Wide RF and IF Bandwidths. The WJ-M12, M12A and M19 series covers 2 to 14 GHz with IF response to 5 GHz allowing single step conversion down to S-band. This series also offers $>+7$ dBm power output capability and superior spurious performance.

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Another one of the W-J mixer family's exciting features is computer characterization. Each mixer is computer tested so you receive complete conversion loss and isolation data up to 12.4 GHz for each unit purchased.

For more information on Watkins-Johnson's computer characterized microwave mixers, and a copy of our Microwave Mixer Catalog, call the Watkins-Johnson Field Sales Office in your area or W-J Applications Engineering in Palo Alto, California at (415) 493-4141, ext. 638.

Pictured is an actual computer characterization print-out for a WJ-M12A Microwave Mixer.



Watkins-Johnson—U.S.A.: 3333 Hillview Ave., Palo Alto, CA 94304 • (415) 493-4141 • TWX: 910-373-1253 • Telex: 34-8415 • Cable: WJPLA • 700 Quince Orchard Rd., Gaithersburg, MD 20760 • (301) 948-7550 • TWX: 710-828-0546 • Telex: 89-8402 • Cable: WJCEI • **United Kingdom:** Shirley Ave., Windsor, Berkshire SL4 5JU, England • Tel: Windsor 69241 • Cable: WJUKW-WINDSOR • Telex: 847578 • **West Germany:** 8033 Planegg, Muenchenerstr. 17 • Tel: (089) 859-9441 • Cable: WJDBM-Muenchen • Telex: 529401 • **Italy:** Piazza G. Marconi, 25 00144 Roma-EUR • Tel: 59 45 54 • Cable: WJROM-ROMA • Telex: 60117

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INFORMATION RETRIEVAL NUMBER 13

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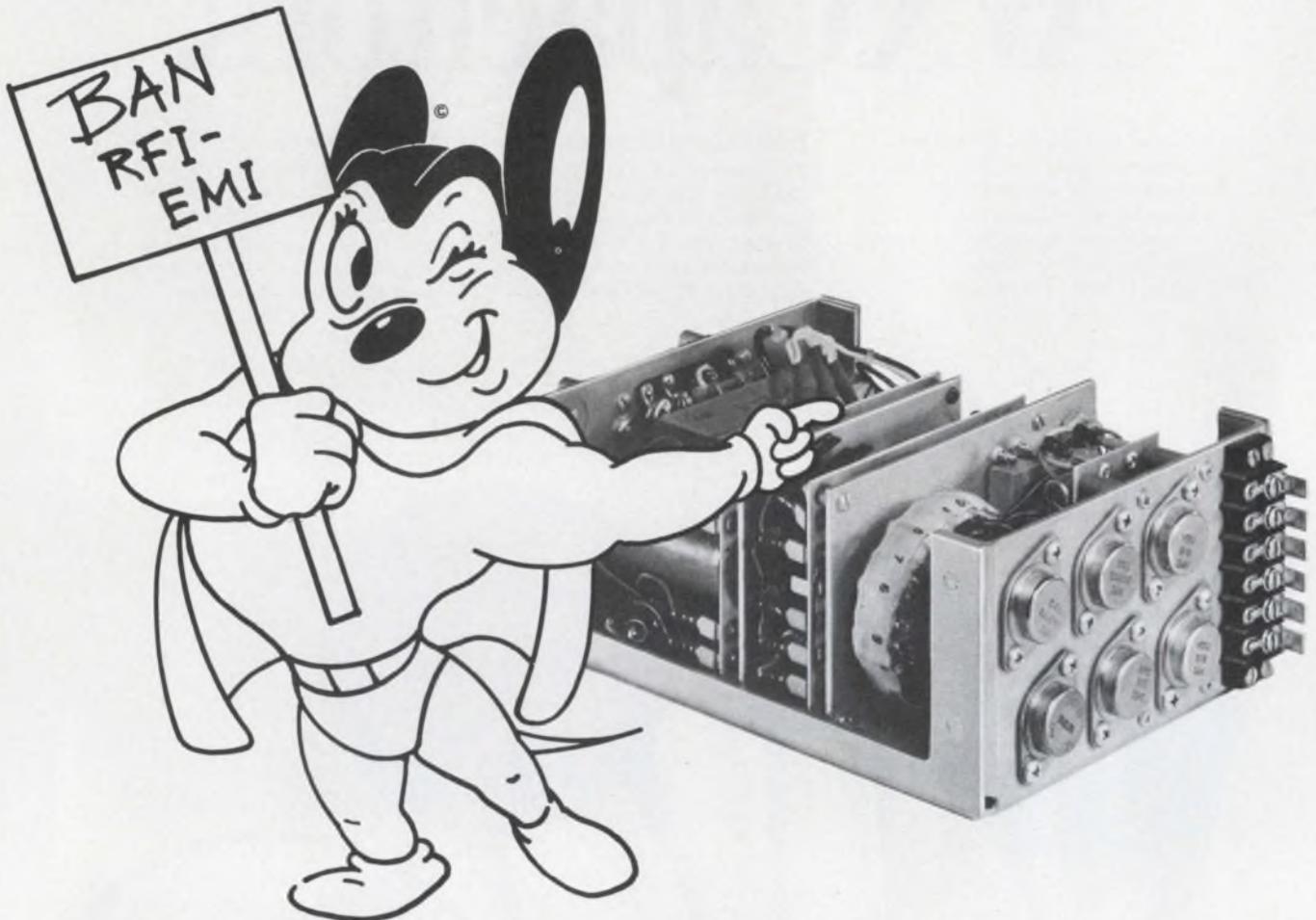
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INFORMATION RETRIEVAL NUMBER 16

ELECTRONIC DESIGN 20, September 27, 1974

Inside AVX Filters.

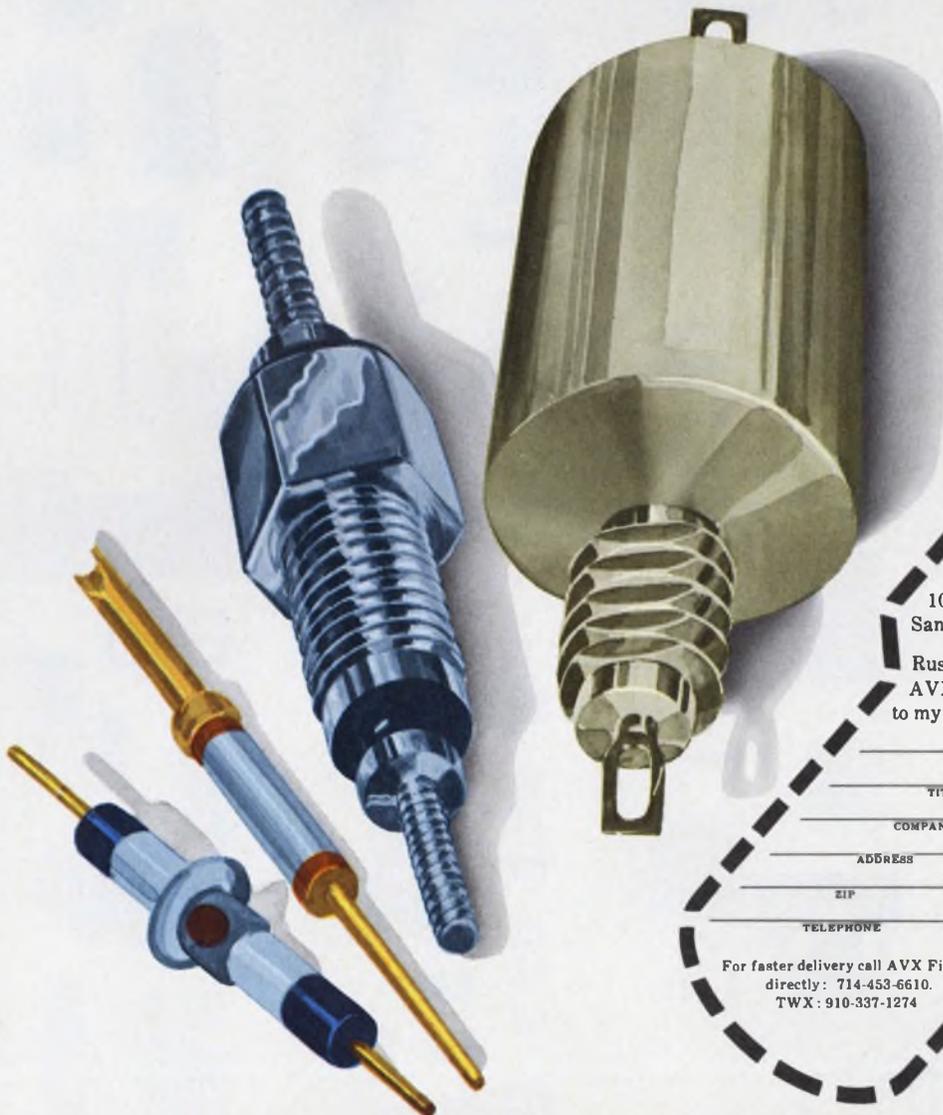
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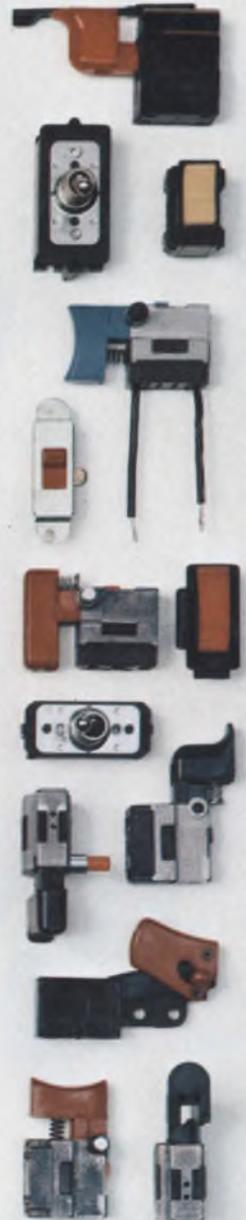
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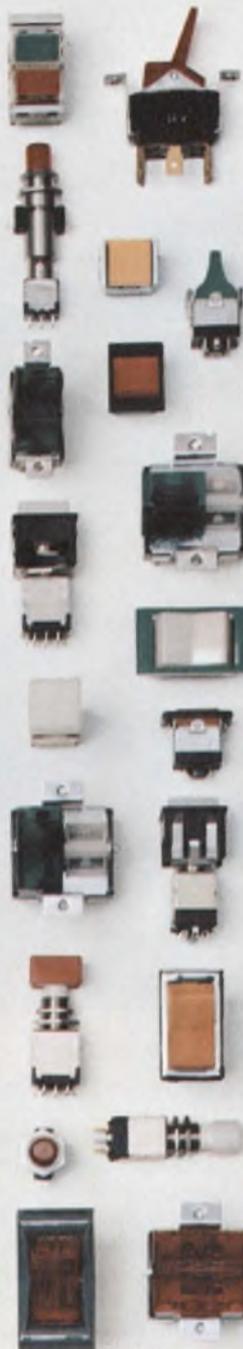


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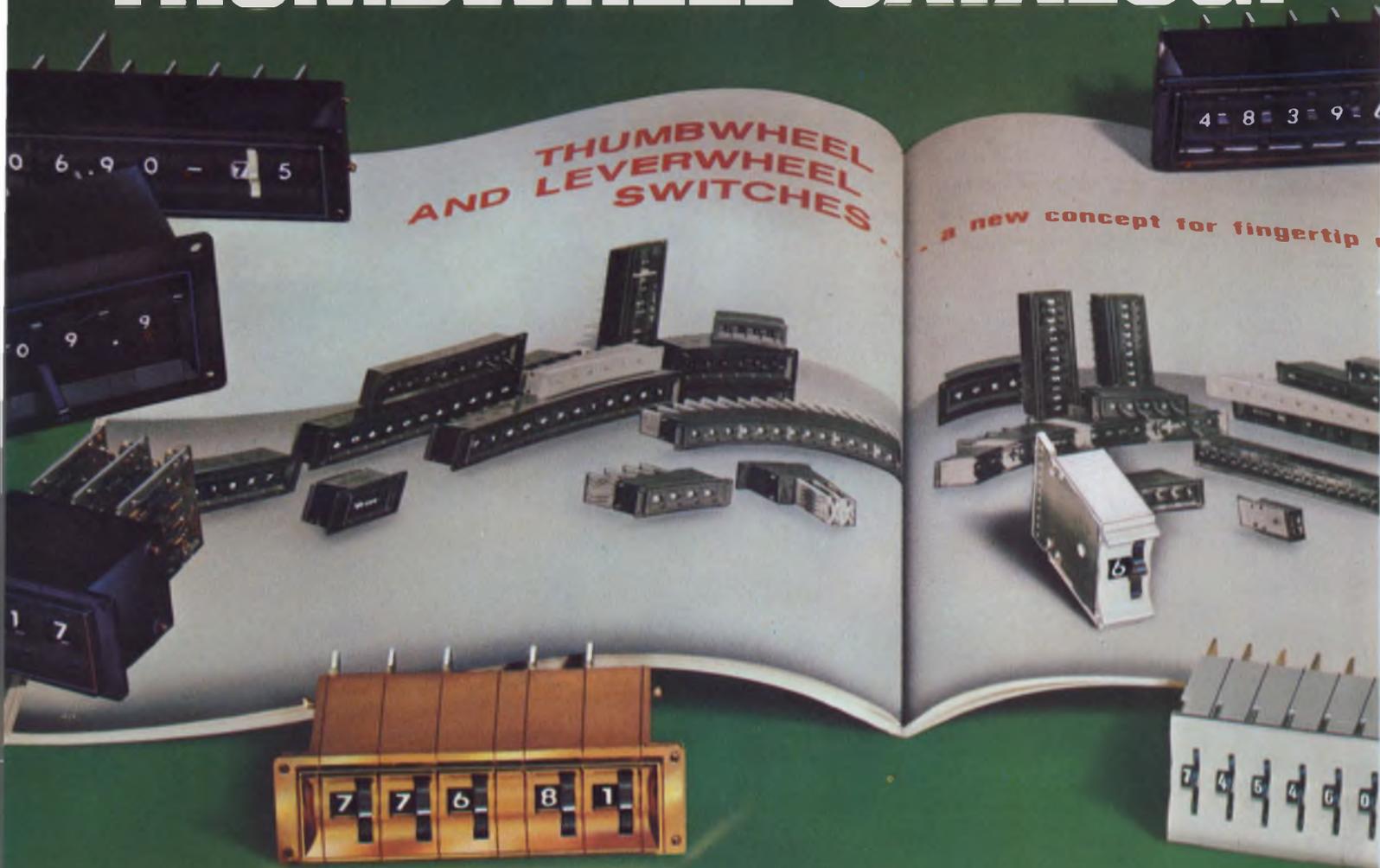
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INFORMATION RETRIEVAL NUMBER 20

Parley will offer solutions to defects in hybrid circuits

Solutions will be offered at a symposium in Boston to one of the most serious problems in hybrid microcircuits—the degradation or destruction of microchip semiconductors due to attack by chemicals in the epoxy adhesives that bond the chips to the substrate.

Thomas Gillis, engineering manager of microelectronics at Raytheon, Quincy, Mass., is technical program chairman of the meeting, which is being sponsored by the International Society for Hybrid Microelectronics. The meeting will be held Oct. 21 to 23 at the Sheraton Boston Hotel.

“Like it or not,” Gillis says, “two years from now, most hybrids will be fabricated with epoxies.”

To avoid disastrous consequences, he warns, “you’ve got to be exceptionally careful in selecting epoxies and in applying them.”

“We’ve done much work on hybrids at Raytheon,” Gillis says, “and we’ve concluded that 50% of the epoxies that are currently available attack aluminum. But aluminum is used to make contacts on the semiconductor chips. So if the corrosive type of epoxy gets on the chip, it eats away the aluminum pattern.”

In Session 6, “Polymers and Hybrids,” successful approaches and continuing problems in the use of epoxy for hybrid fabrication will be described by experts from six companies—Applied Technology, Raytheon, Autonetics, Airborne Instruments Laboratory, Analog Devices and IBM.

Substantial progress has also been made in reducing the size of microelectronic hybrid packages, according to Dr. Robert Mandal, staff scientist at Applied Technology, Sunnyvale, Calif., Mandal, chairman of Session 7, “Hybrid Packaging,” points out:

“Today we are doing a lot more

in a smaller volume. For example, we’re packaging very high-density memory modules.”

Important factors in size reduction, Mandal explains, have been improved methods of heat sinking, both within the microcircuit and between the hybrid package and environment. Greater reliability and reduced circuit cost result, he notes.

But Mandal points out that a big problem remains: the elimination of interconnections.

“Even though integrated microcircuit techniques have reduced interconnections for a given circuit by, say, 80%,” he says “we still have wirebonded interconnections within the hybrid package. And that’s something we want to get rid of.”

Computer control of hybrid microcircuit design and fabrication has produced improved circuit performance notes Daniel D. Zimmerman, physicist on the senior professional staff at Johns Hopkins University Applied Physics Laboratories, Silver Spring, Md.

Zimmerman, chairman of Session 5, “New Applications for Hybrids,” points out that computer-controlled laser-trimming techniques have produced a new concept in hybrid circuit production, called functional trimming.

“You no longer trim a resistor to a given value, but you trim the circuit to a desired transfer function,” he notes.

“This provides improved operating characteristics and is very cost-effective.”

Built-in microprocessor available in scope

Automatic measurement and display of time interval, frequency, dc voltage, instantaneous ac volt-

age and relative amplitude expressed in percent are available on the first microprocessor-controlled oscilloscope.

Called the 1722A by its manufacturer, Hewlett-Packard in Colorado Springs, Colo., the scope is a dual-channel, 275-MHz laboratory model with 6 × 10-cm display. It sells for \$4500 without probes or other accessories.

“The microprocessor is the same chip set that runs the HP-35 hand-held calculator,” according to Walt Fisher, project manager. American Micro Systems [Santa Clara, Calif.] supplied the chips for the first production run. Mostek [Carrollton, Tex.] is an alternate source for the chips.”

Calculated results are displayed on a 3-1/2-digit LED readout with scientific notation. The exponents are always given as -3, -6 or -9. When used for time-interval measurements, these correspond to ms, μ s and ns.

To make the time-interval measurement easier a dual delayed sweep function has been incorporated. This is a dual comparator that fires the sweep at one voltage and their again at another. The start and stop points of the desired time interval then appear as intensified markers on the scope face. A simple switch inverts a time-interval measurement to give a direct frequency readout on the LED display.

The new scope is expected to be the first of a family. Later products will probably have such features as user programmability and digital storage of a waveform for future reference. There might even be an HP-35 keyboard on the scope control panel.

Amplifier tube claims ‘firsts’ in performance

A new jumbo low-light-level amplifier tubes for Air Force night photo reconnaissance missions is reported to set a number of firsts in performance. Developed for the Air Force Avionics Laboratory at Wright-Patterson AF Base, Ohio, by ITT, Fort Wayne, Ind., the single-stage, image-amplifier tube is focused electromagnetically and has an active aperture, or useful diameter, of 162 mm (6.4 in.).

A prime advantage of the tube, according to Frederick Farinet, Air Force project engineer, is that it permits use of a camera with standard 5-in. film and gives full area coverage. Previously, with the next largest tube (144 mm, or 365 in.), the image was clipped at the corners.

Compared with the 144-mm device, the 162 mm is said to have substantially improved optoelectronic characteristics:

- Cathode sensitivity is high across the operating band of 700 to 980 nanometer (nm). Response at 850 nm is 20 mA/W compared with 3 mA/W for the smaller device.

- Stage gain is 28 W/W, contrasted with 5 W/W for the 144-mm unit.

- Limiting resolution is 75 line-pairs per mm per stage, compared with 60 for the smaller tube.

- The tube can be efficiently gated on and off in less than 10 μ s, compared with 10,000 μ s required by the 144 mm tube. This gating feature permits a substantial reduction in atmospheric back scatter from the aircrafts gated light sources. Although 10- μ s gating has been achieved with photo multipliers, it is the first time that such rapid gating has been produced with large-image tubes.

- For mapping purposes, a reticle is incorporated in the acthode substrate. Earlier tubes did not have this feature.

Farinet points out that a major step has been taken toward achieving essentially uniform sensitivity over the full cathode area of the tube. The loss of sensitivity from the center outwards to the edge of such tube cathodes has been a problem for years.

Because of the increased gain of the new tube, an aircraft taking covert night pictures will be able to operate at higher altitudes than heretofore.

C-Band landing system adopted by the FAA

An interim microwave landing system, compatible with the ILS avionics currently installed in aircraft, has been adopted by the Federal Aviation Administration.

Developed by Tull Aviation

Corp., Armonk, N.Y., the system operates at 5 GHz in the C band and has separate transmitters and scanning antennas for the localizer and the glide-slope functions. And it requires only a receiver-converter to change the pulsed C-band radiation to standard vhf/uhf ILS signals. The output of the converter is fed to conventional ILS receivers.

After competitive evaluation and flight tests were made with microwave equipment produced by Singer-Kearfott and Boeing, the Tull system was selected as the only one to pass all Category I landing-systems specifications.

The system is intended for use at locations where the standard ILS has reflection and interference problems, or where the needs for the low-approach service would be better met by installation of the newer system instead of the older ILS.

Under FAA contract terms, Tull will license companies wishing to build its system. Requests for data should be made directly to Tull.

An FAA spokesman says that limited application of the system will not affect the national microwave landing-system program. The FAA plans to amend its aviation regulations to provide financial aid under the Airport Development Program for the new installations.

Walter Barry, Tull's manager of installations, estimates that the cost will be about \$110,000 to \$125,000. Present costs of the receiver converter for the aircraft are projected at \$3000.

FM stations reducing noise the Dolby way

"Dolbyize" has started in the wake of a public notice by the Federal Communications Commission that it has no objection to broadcasters combining the Dolby B-type noise-reduction system with a 25- μ s pre-emphasis curve.

The noise-reduction system, invented by Dr. Ray Dolby of Dolby Laboratories, New York City, selectively boosts the amplitude of the high-frequency components of audio material during low-volume music passages. This coded material is then recorded or broadcast. Upon playback or reception, the

process is reversed to return the audio response to normal. To the listener, the high-frequency noise produced by the tape or receiver is attenuated by about 10 dB during the quiet passages, where it is most objectionable.

A number of commercial radio stations, including WQXR in New York, have been broadcasting Dolbyized programs for some time. The problem that has prevented wider use is that a listener without decoding equipment hears the program is "overly bright" because of the boosting of the audio high frequencies.

By a change in the audio pre-emphasis (boost of high frequencies) in a transmitter from the standard 75 μ s to a less-severe 25 μ s, this problem is solved. Receivers of tuners without Dolby decoding hear the program with its normal proportion of high frequencies. Dolby-decoder-equipped receivers, provided with 25- μ s de-emphasis, receive the program normally, with a 10-dB better signal-to-noise ratio.

"So far we have delivered over 20 of the Type 324 encoder units incorporating the 25- μ s pre-emphasis," says Morely Kahn, Dolby's vice president and U.S. manager.

Secure messages sent via phone by terminal

Through electronics the Air Force may turn a secretary into a communications center, provided she has, among her other assets, a security clearance.

The tool is a compact message terminal that contains a built-in electric typewriter connected to a cassette tape drive, which connects to a secure cryptographic processor in a safe. Incoming messages are automatically recorded, while outgoing messages are composed on the typewriter, transferred into digital format through the tape unit, processed by the cryptographic device and transmitted over ordinary telephone lines.

The Air Force envisions a number of terminals on each base that would tie into a computer-controlled automatic switch developed by the Air Force Communications Service's Communications Computer Programming Center.

What's going in... in mainframes?

MOSTEK's 16-pin 4K RAM.

Your mainframe memory system... MOSTEK's 16-pin 4K RAM. A perfect match. Because MOSTEK's 4K RAM gives you the performance, board density and ease of use your system demands. Plus it's available *now*.

Compare performance at the system level. MOSTEK's low capacitance, TTL compatible clocks, combined with superior output drive capability, provide the access time your system requires. Alternate 22-pin designs require high capacitance clocks and additional output buffering, causing system speeds to be lower. MOSTEK's 16-pin multiplexed design reduces the number

of address buffers required without affecting high speed access time. For small peripheral memory arrays the cost savings are substantial — even more so in the case of large mainframe arrays. So look to MOSTEK to meet the high speed requirements of your system.

Want board density? Of course. All memory users — from peripheral and minicomputer manufacturers to the big mainframe people — appreciate the increased density offered by MOSTEK's 16-pin design. (A 50% savings in memory board size over 22-pin alternates.) The result is a more compact, cost effective system.

Interested in ease of use? Again, compare the advantages of MOSTEK's MK4096. Readily available automatic insertion equipment can be used in

board assembly. Voltage pins are on the corners to simplify PCB layout. All inputs including clocks are directly TTL compatible with low capacitance. And the circuit is extremely tolerant of noisy system environments.

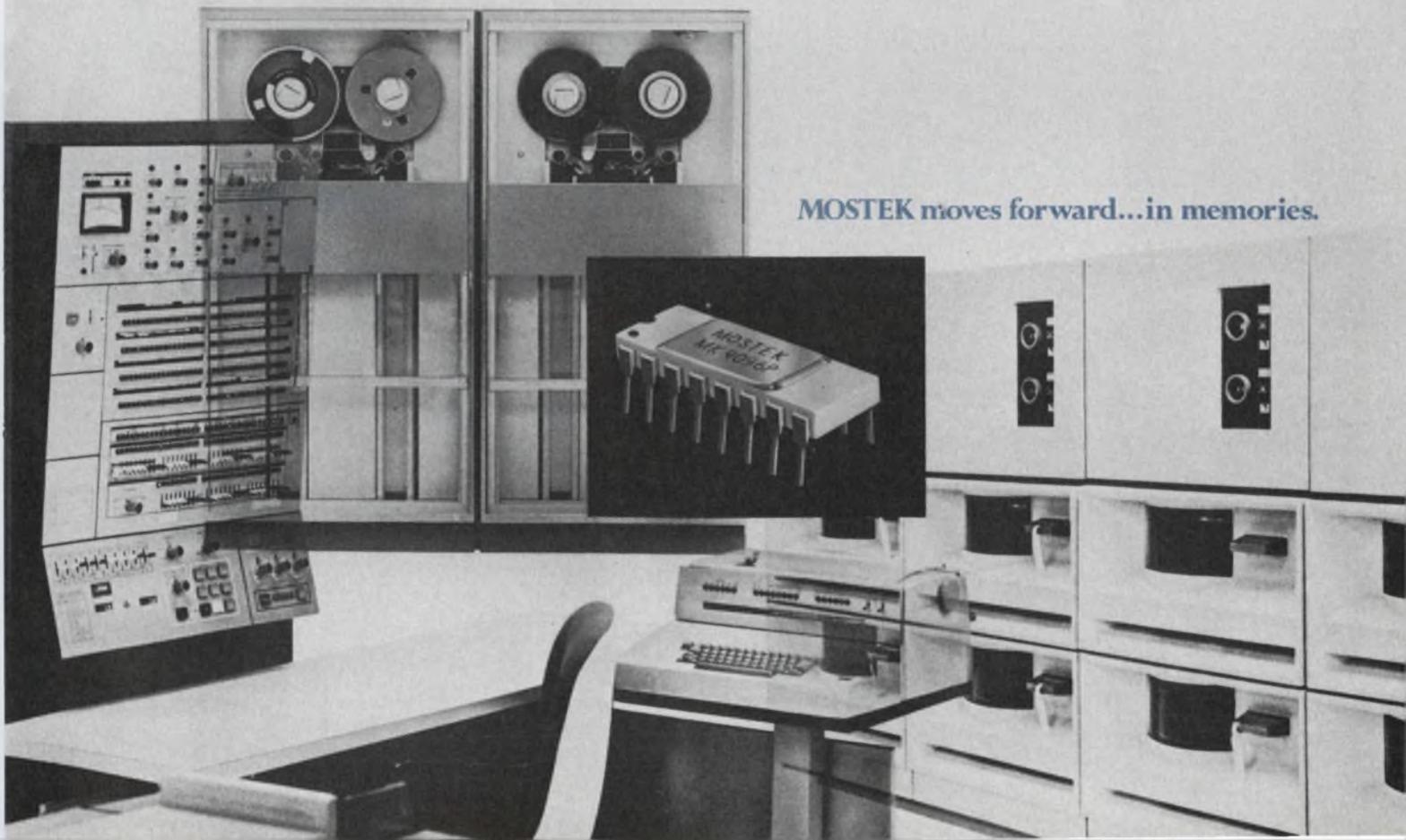
Your mainframe, minicomputer or peripheral memory and MOSTEK's 4K RAM. That's what's *going in*.

Want more details? Call your local MOSTEK distributor or representative or contact MOSTEK, 1215 West Crosby Road, Carrollton, Texas 75006, (214) 242-0444. In Europe contact MOSTEK GmbH, TALSTR. 172, 7024 Bernhausen, West Germany, Tel. 798038.

MOSTEK

INFORMATION RETRIEVAL NUMBER 24

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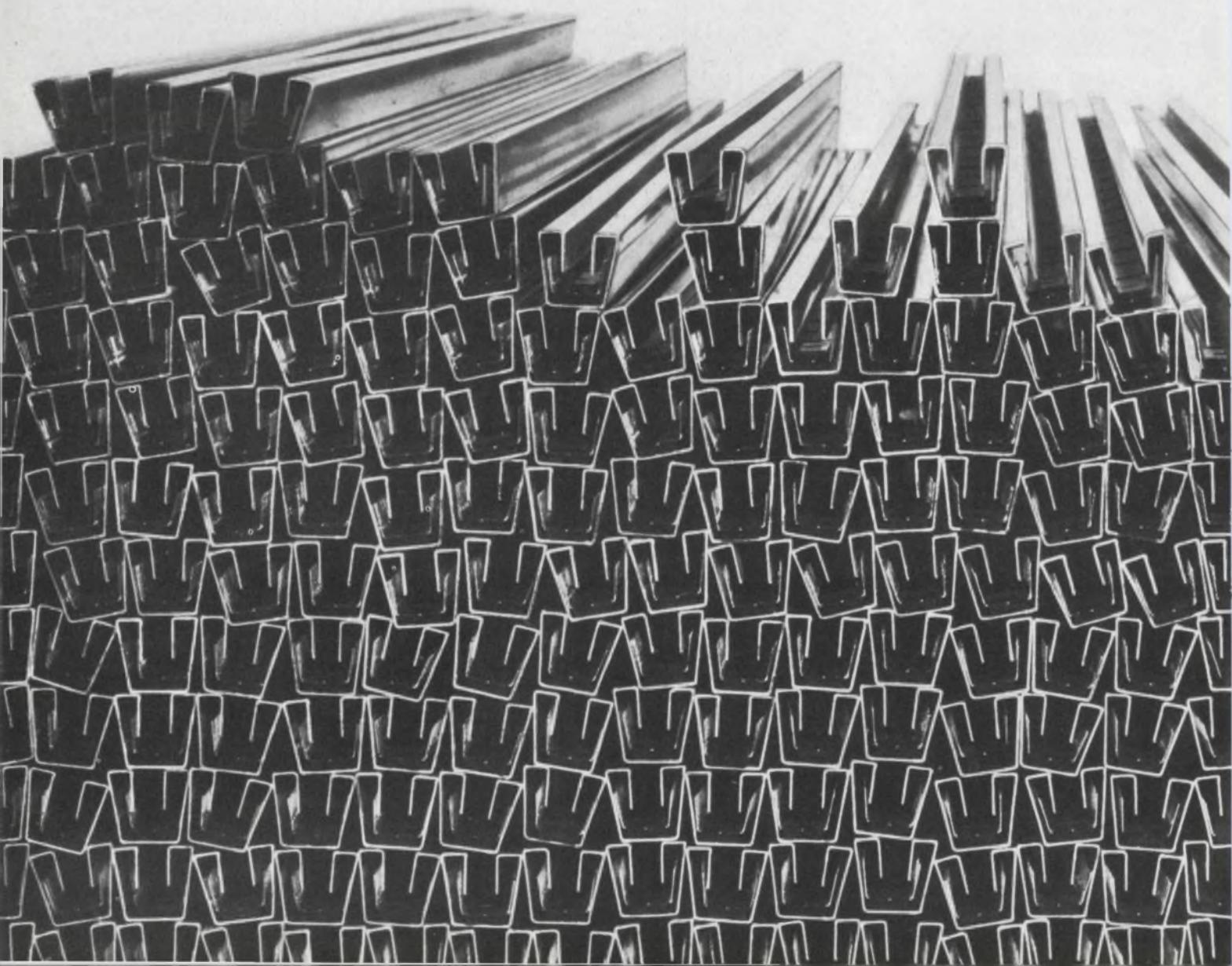
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other guys' standard CMOS devices. We're also turning out a lot of our own originals.

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brand new three-inch wafer facility; backed by ion implantation, automatic wire bonding, computer-controlled testing. That all adds up to high quality and high yield. Consistent availability. In volume.

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INFORMATION RETRIEVAL NUMBER 25



A little-known source of power, the thermal cell, bids for attention

"Now that some of the military restrictions have been eased and more civilian designers are finding out what thermal batteries can do, the demand could lead to some very effective low-cost designs."

This is the view of Frederick Tepper, vice president and general manager of Catalyst Research Corp., Baltimore, and to back it, he points to advances that have been made in the thermal battery construction.

Representing a unique source of electrical energy, thermal cells can produce hundreds of watts of peak power in packages weighing less than a pound. However, once activated, they produce power for only a short time, ranging from milliseconds to perhaps half an hour for different versions.

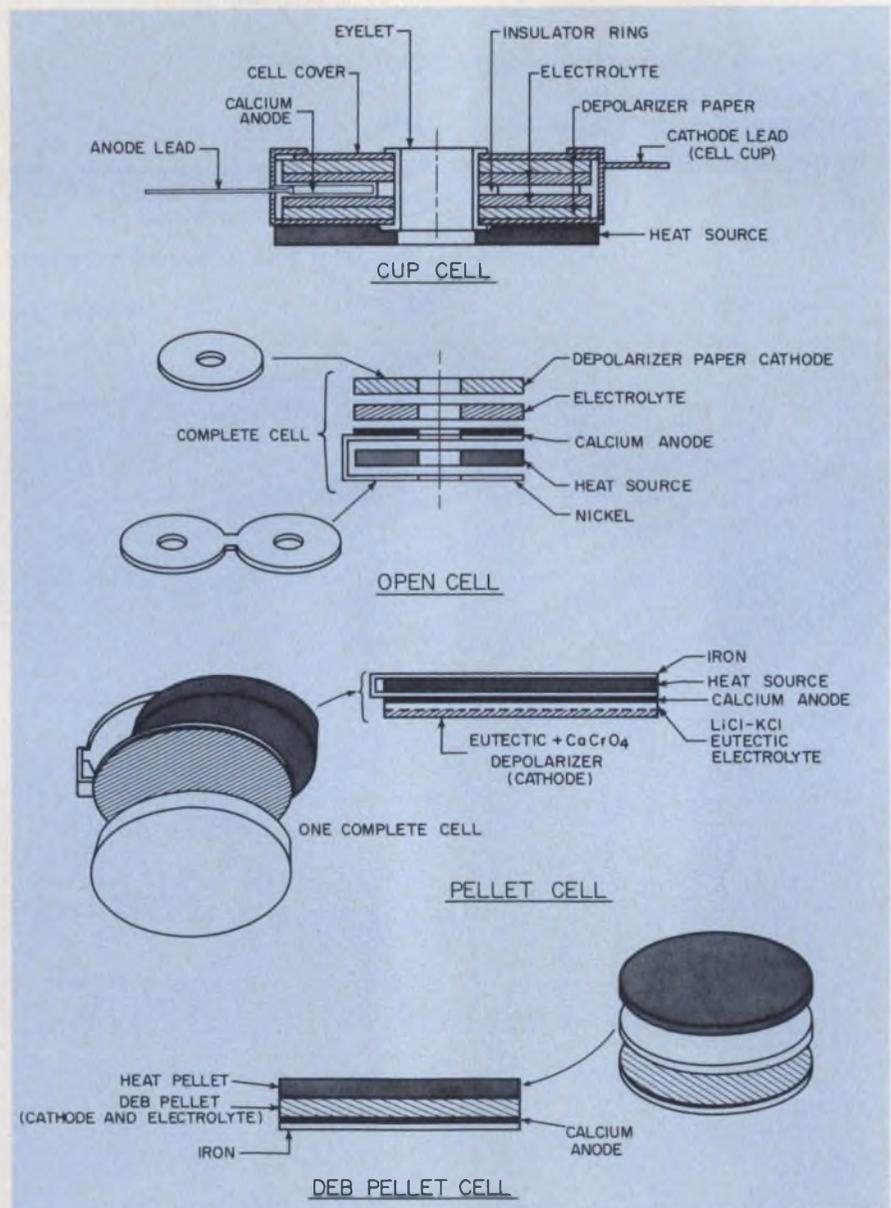
Can be stored for years

Thermal batteries have been used to power military ordnance, such as time fuses and missile electronics, since their invention in the late 1940s. But until fairly recently they have remained virtually unknown to designers of civilian electronics.

Present military thermal batteries have proved reliable sources of power for equipment that may be stored for years and then used at a moment's notice.

Not only is the battery ideal for weapons systems but for critical emergency equipment, such as aircraft ejection systems, emergency transmitters and signaling devices.

The military batteries range in



Most thermal battery cells are made of layers of a gasless pyrotechnic heat source, anhydrous metal salt electrolyte, calcium anode and metallic oxide, sulfide or chromite cathode. The older designs, cup and open cells, are made of paper-like layers of the materials, while the newer cells use pellets, which are mechanically stronger and easier to assemble. In the DEB (depolarizer, electrolyte, binder) cell, the cathode and electrolyte are combined in a single homogeneous pellet.

Northe K. Osbrink
Western Editor

Characteristics of some typical thermal batteries

| Cell type | Volume (cc) | Weight (g) | Nominal voltage | Current (A) | Peak power (W) | Average life (sec) |
|------------------------|-------------|------------|-----------------|-------------|----------------|--------------------|
| Cup | 450 | 850 | 7* | 5.8 | 41 | 70 |
| | | | 126 | 0.47 | 60 | 70 |
| Open cell | 100 | 385 | 50** | 0.36 | 15 | 70-100 |
| Open cell | 44 | 148 | 18 | 26 | 462 | 1.2 |
| Open cell | 1 | 5.5 | 10† | 5.0 | 50 | 0.15† |
| High voltage array†† | 81 | 225 | 203 | .02 | 4 | 45 |
| | | | 1.7 | .32 | 0.5 | 45 |
| Heat-DEB | 2.9 | 7.5 | 20 | 0.045 | 0.9 | 25 |
| Heat-DEB | 105 | 307 | 28 | 1.2 | 34 | 150 |
| | | | 28 | or 2.5 | 75 | 60 |
| Heat-DEB | 592 | 1560 | 28 | 2.8 | 78 | 800 |
| Activated heat-DEB | 83 | 250 | 45 | 6.0 | 385 | 25 |
| Advanced pellet design | 570 | 1485 | 28 | 5.5 | 158 | 720 |

*Two-section battery.

**Three sections totalling 58 V.

†Open circuit voltage is 33 V. Thermal life (light load, 28 V operation) is 2-3 seconds.

††Multisection battery with power and high voltage sections.

price from \$2 for some high-production units to \$200 or more, but civilian versions could be far less expensive, Tepper says.

"So far there have been no large production runs or automated assembly," he notes. "Even though over 10 million thermal batteries have been produced. We have been, in effect, job-shopping them and making a few for each contract—never more than two or three thousand a month."

The thermal cell is a primary cell, producing an electric current by the chemical action of oxidizing a metallic anode and reducing a cathode in the presence of an electrolyte. Unlike the more common primary cells—such as zinc-carbon or mercury, which are actively producing current immediately after assembly—the thermal cell is a reserve cell. The chemical activity does not start until an internal source of heat is activated. So the shelf life is virtually infinite.

The thermal battery uses an anhydrous salt electrolyte that operates at high temperature—usually molten. The high temperature is provided by inclusion of a non-gas-producing chemical heat source within the hermetically sealed battery case. Activating the battery requires that the heat source be ignited—with an electrical squib, firing pin or fuse.

According to Tepper, the most common thermal cells use a molten lithium and potassium chloride mixture for the electrolyte with a calcium metal anode. The active cathode material, called the depolarizer, may be oxides of iron, vanadium or tungsten and sulfides of copper or calcium chromate. Such cells require a temperature of 300 to 500 C to melt the electrolyte and to activate the battery.

The first practical thermal batteries were built with the cup-cell technique, and many current production batteries use the same construction. The active materials are contained in a metal cup, which also serves as the cathode connection. In the cup cell—and a later variation, the open cell, which eliminates the containing cup—the source of heat is a compound of zirconium and barium chromate manufactured in a paper-like sheet. It tends to be fragile and susceptible to accidental ignition.

Pellet cells replace paper

The paper cells have largely been superseded by pellet cells. In these the heat source—electrolyte, anode and cathode—is manufactured in the form of round pellets, which are comparatively sturdy and easy to handle. The electrolyte—also barium-zirconium chromates—is mixed with a binder like kaolin,

which causes it to maintain its shape during the operating cycle.

In 1959, Catalyst Research Corp. and Sandia Laboratories, Albuquerque, N.M., developed independently two versions of a new type of thermal cell. Known as the DEB (depolarizer, electrolyte, binder), the cell is built like other pellet cells. But the depolarizer (cathode) and electrolyte are combined into a single homogeneous pellet, held together with a binder.

Tepper notes: "You theoretically could get some cell discharge, since the anode is placed directly next to the cathode—an ion short. However, in operation, the molten blobs of depolarizer are surrounded by extremely thin layers of pure electrolyte. So, in practice, there is no loss of efficiency in using the DEB pellet."

By combining the DEB construction with an improved heat source (iron-potassium perchlorate), the "activated heat DEB" cell is capable of longer operating life and more stable power production.

During combustion, Tepper explains, the perchlorate gives up its oxygen to the iron, leaving a pellet of potassium chloride when combustion is finished.

"The chloride acts as a thermal sink, melting and absorbing heat during combustion and giving it up as the cell cools," Tepper explains. ■■

Ulf direction finder for miners passes tests at 1500-ft depth

An ultra-low-frequency electromagnetic direction-finding system that can pinpoint a transmitter 1500 ft. below the earth's surface is ready for operational use after more than three years of development and field-testing. It will be used to find miners who are trapped in underground disasters.

The system comprises a miner's keyed cw transmitter connected to a large loop antenna, which radiates at a discrete frequency in the 900-to-3100-Hz band. Above-ground receivers, with direction-finding loops, are designed to be carried by men on foot and by flying helicopter.

The system, developed by West-

inghouse for the United States Bureau of Mines, has proved capable of locating the position of underground transmitters to within one foot for shallow mines of 200-to-350-foot depths (see table). For a worst-case test, Westinghouse chose the Geneva Coal Mine, near Dragerton, Utah.

Here, the ground has relatively high conductivity and the overburden—the depth of the earth above the mine passageways—is deep. At the shallowest overburden location of 1150 ft., the signal from the mine could be picked up by the helicopter as it flew at 100 ft. The horizontal detection range was about 4000 ft. Signals were also detected by the helicopter at depths up to 1500 ft., but the horizontal detection range was limited to about 1000 ft.

The helicopter was used for quick scans over the mine area, to determine the presence or absence of transmitted signals in the mine and their approximate location. The man-pack receivers were used to zero in on the signal source.

The direction-finding system is being refined by Collins Radio Co. in light of the tests, and Collins will build the production models.

The original design of the ulf system made use of the fact that miners equipped with tone transmitters driving a large loop antenna could generate a vertical magnetic dipole moment that could be located by measurement of the magnetic fields on the surface. Directly above the dipole or source, is a null that provides the location.

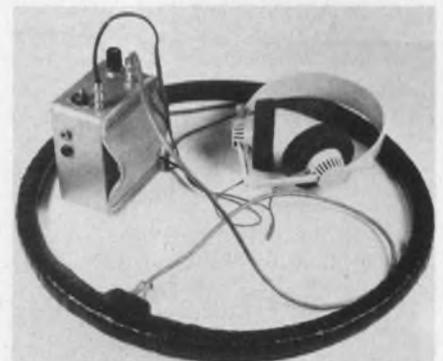
The accuracy of the location depends upon the depth of the null, which is a function of signal level and background noise. Accuracy also depends on the orientation of the axis of the underground loop, as well as on the slope of the terrain on the surface. However, the tests have demonstrated that correction factors can reduce the slope error to a small one.

(continued on p. 34)

Jim McDermott
Eastern Editor



Underground field strength measurements are being made by Westinghouse engineers in the Eagle Coal Mine in Colorado. Evaluation of the penetration depth of ulf radio transmissions was aided by these measurements.



This prototype ulf receiver, by Westinghouse, uses a direction-finding loop to find the null in the pattern being radiated by underground transmitters.

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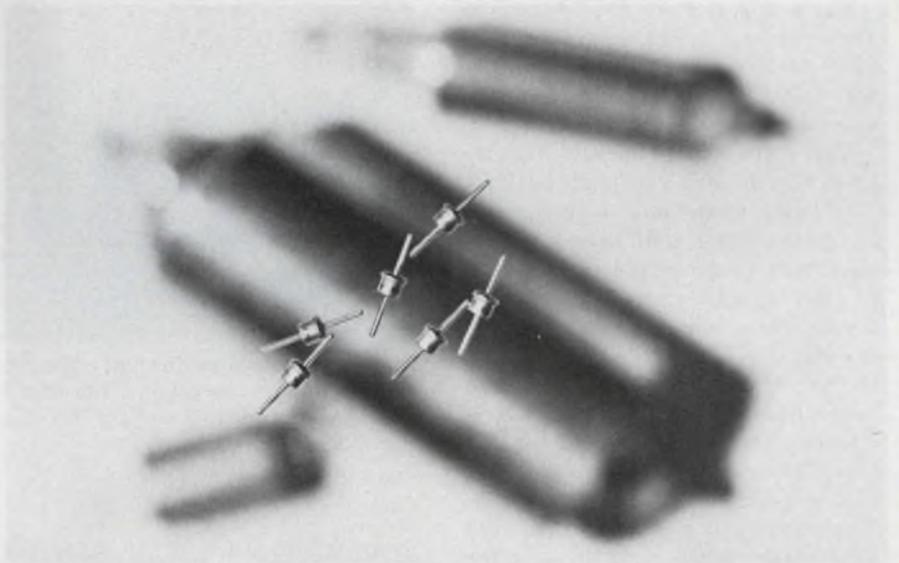
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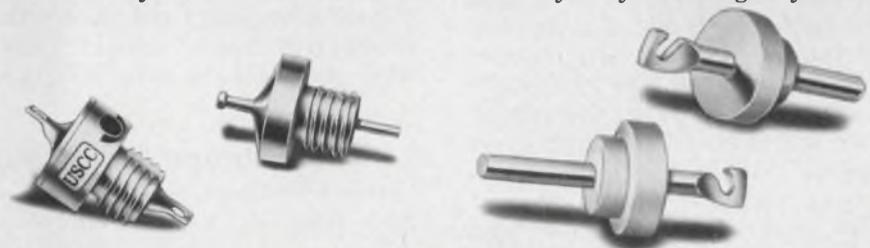
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INFORMATION RETRIEVAL NUMBER 26

Two types of prototype receivers were developed by Westinghouse: a portable, single-channel receiver with an aircore loop antenna tuned to one particular frequency (see photo) and a multichannel helicopter receiver with a broadband loop and preamplifier.

The Westinghouse manpack receiver is a tuned audio amplifier employing four bandpass filters and a sharply tuned, microelectronic piezoelectric tuning-fork filter with a Q of 500. The overall receiver bandpass is 3 ± 2 Hz. Out-of-band interference ± 300 Hz away from the center frequency is attenuated about 70 dB.

Each receiver is fixed-tuned to frequencies of 922.5, 982.5, 1700, 1900, 2500 and 2900. The helicopter unit can receive all six simultaneously.

The transmitters operate directly from the miner's cap lamp battery. The basic audio tone is generated by an oscillator that uses a microelectronic piezoelectric tuning fork as the frequency-generating element. The oscillator's accuracy is ± 2 -Hz.

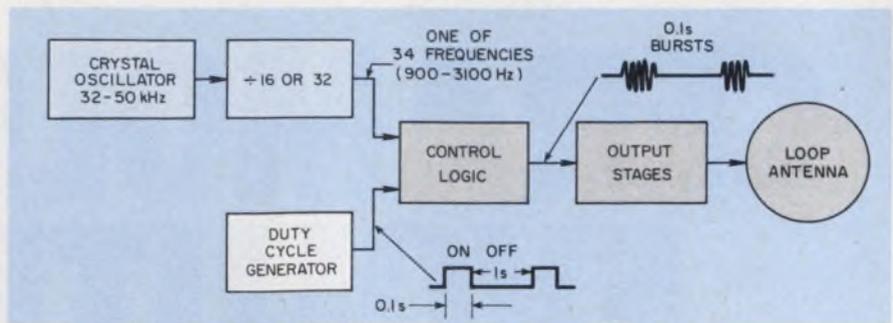
The transmitter output is a gated burst of the tuning-fork frequency—on for 0.2 s and off for 2 s. The output is 8 amperes, peak, into a $0.4\text{-}\Omega$ resistive load at 70 F. The antenna is a loop of wire that the miner winds around a mine pillar.

It's possible to key the transmitter manually for coded communication, or it can be set for automatic burst operation.

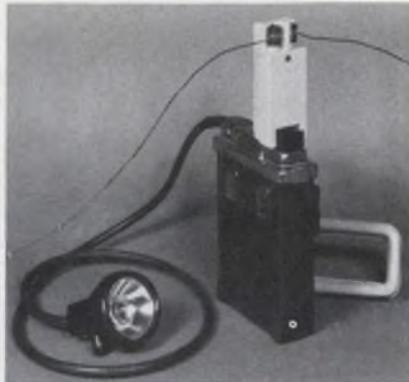
Both the transmitter and receiver have been miniaturized for greater portability in the Collins production version. The original designs used microelectronic tuning forks in both the transmitter and receiver. But in the production transmitters the tuning-fork crystal has been replaced by a crystal oscillator in the 32-to-50-kHz range. This oscillator uses electronic-watch crystals.

The crystal output in the Collins transmitter has been divided down to provide a single basic frequency in the 900-to-3100-Hz range. Different frequencies will be assigned to different sites in a mine.

To reduce battery drain and provide 10 hours of minimum operation after the miner has used his cap lamp for eight hours, the transmitted burst has been reduced



The production u/f Collins transmitter uses a watch-crystal oscillator that is counted down to give discrete frequencies in the basic audio range suitable for underground communication. Loop is spread out in mine.



The u/f transmitter for underground use by miners, by Collins, fits into a clip on top of the battery for the miner's lamp.

from 0.2 to 0.1 s. This called for an increase in receiver bandwidth from about 2 Hz to 10.

Another battery-saving feature has been built into the transmitter, according to Dean Anderson, head of mine electronic product design for Collins. As the battery gets weaker, he points out, the voltage drops more and more with transmitted bursts. But it rises when the transmissions stop. To

keep the crystal oscillator in operation despite the drop in voltage, a capacitor is charged by the battery during the off portion of the transmitter cycle. During the on portion the oscillator runs off of the capacitor.

The transmitted output power into a typical loop, Anderson says, is about 10 W, average, during the sine-wave burst.

To eliminate 60-Hz interference in the receiver, the transmitter frequencies fall between the 60-Hz harmonics.

In field tests the microelectronic tuning-fork filter in the receiver proved to be microphonic. This noise interfered with reception by the manpack receiver. In Collins version, the fork has been eliminated. And the circuit, instead of being a tuned audio amplifier, has been redesigned as a superhet with an intermediate frequency of 50 kHz.

"We're translating the input signal upwards to 50 kHz and putting it through a 50 kHz i-f crystal filter to give us the required 10-to-15-Hz bandwidth," Anderson reports. ■■

Prototype u/f locator system performance

| Mine | Overburden thickness, ft | Hill slope, deg | Location accuracy, ft | |
|----------------------------|--------------------------|-----------------|-----------------------|--------------------------|
| | | | Uncorrected | Corrected for hill slope |
| Inexco #1 Jamestown, Col. | 350 | 23.5 | 43 | 1 |
| Somerset Paonia, Col. | 500 | 15 | 23 | 5.6 |
| Copper Queen Bisbee, Ariz. | 200 | 15 | 5 | 1 |
| Geneva Dragerton, Utah | 130 to 200 | 22 | 130 to 300 | — |

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INFORMATION RETRIEVAL NUMBER 27

Hand-held gas detector promised, with 1-part-per-billion sensitivity

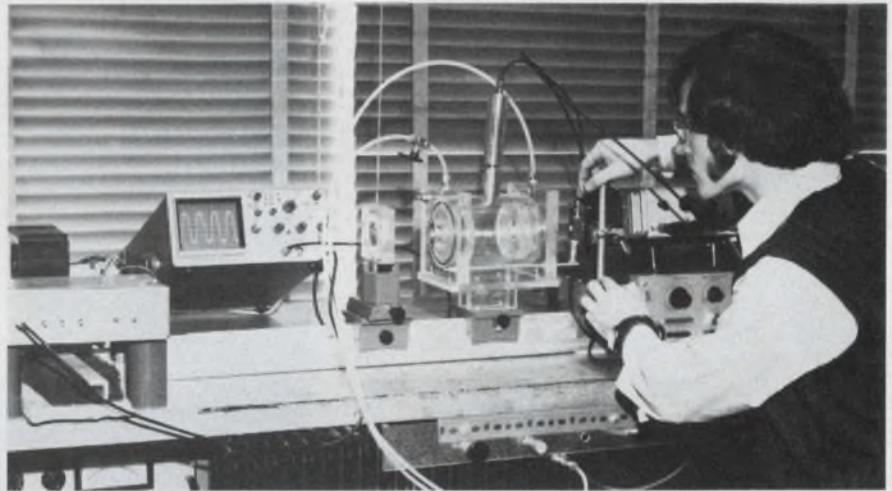
A gas-detecting system that could pinpoint air pollutants in concentrations of one part per billion may be one answer to a demand for hand-portable detectors to sniff out explosives and dope in packages and luggage at transportation and postal terminals. Present equipment used for this purpose requires a truck to carry it.

The new detector, developed at Massachusetts Institute of Technology under the sponsorship of the Advanced Research Projects Agency, is an opto-acoustic system. It produces acoustic signals by the rapid expansion of a pollutant gas as it absorbs energy from the chopped beam of a laser (Fig. 1). The laser wavelength is chosen to coincide with an absorption line of the gas.

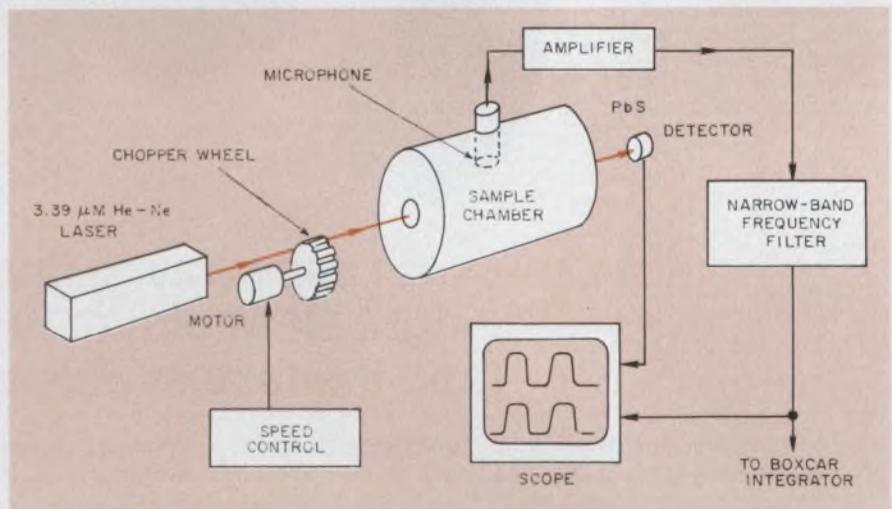
The prototype model has a sensitivity of 200 ppb, but one of its developers, Roger Kamm, research assistant at MIT, says that with relatively simple modifications, the system would be able to detect as low as 1 ppb of an air contaminant.

The basic approach is not new, according to Associate Prof. C. Forbes Dewey Jr., who designed the improved version. Earlier systems, he points out, suffered from sensitivity losses, caused by absorption of energy by the test chamber walls and end windows. Consequently the pressure signal that activated a microphone inside the chamber was very low.

These problems were overcome with three new features in the MIT system design. First, the sample cell was designed as a resonant acoustic chamber, so that small pressure fluctuations produced by the chopped laser beam could be amplified by the Q of the acoustic chamber.



Acoustic chamber of MIT gas-detecting system is being aligned with laser beam by Roger Kamm, one of system developers. Output of the photodetector, used for alignment, appears on the screen of the scope. Laser is at left.



The infrared absorption of a gas sample is amplified by the acoustic resonance of the sample chamber. The chopped laser beam, in passing through the chamber, produces sonic pressure changes picked up by the microphone.

The maximum Q of the chamber approached 700, but the gain of the prototype system—due to other losses—was 100 times that of a nonresonant system.

To further enhance the acoustic signal gain, the laser beam was

chopped at the chamber's 3900-Hz resonant frequency. This is in contrast with earlier nonresonant systems using 100-Hz chopping rates.

The resonant-frequency chopping technique overcame another major
(continued on page 38)

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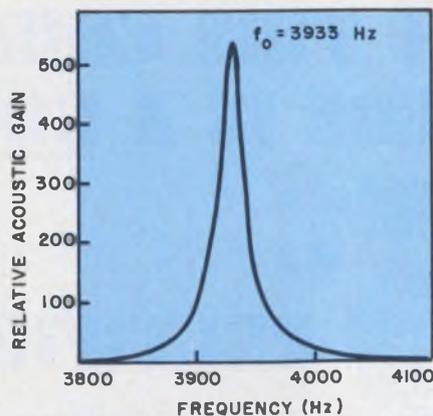


problem of earlier systems—absorption of the laser radiation by the sample cell windows and walls. This produced pressure signals that were identical to those produced by the pollutant being measured.

Dewey points out that because the resonant gas-pressure waves are different from the waves triggered by window and wall absorption, the latter do not interfere with detection of the pollutant signal.

In the prototype experiments, a 2-mm beam from a 0.6-mW, 3.39- μ m helium-neon laser made a single pass through gas in a chamber. The sample chamber was 92 mm long and had a diameter of 102 mm.

To pick up the pressure waves, a 5-mm electret microphone, encapsulated with an impedance-matching FET, was suspended inside the chamber at the end of a small rod. And to increase the amount of energy absorbed by the gas internal concave spherical re-



The measured acoustic gain of the sample chamber approaches 600.

flectors were added to both ends of the chamber.

Dewey estimates that the beam was reflected back and forth some 20 times with this arrangement. The mirror at the laser-entrance end was drilled with a small center hole to permit the beam to pass into the chamber.

Losses with this optical arrangement are about 5%, Dewey re-

ports. But this can be reduced, to 0.2%, he adds, by the use of anti-reflection coatings and the installation of external mirrors.

Experiments with n-butane as the trace absorbing gas have verified the soundness of the system concept, Dewey says. The output was demonstrated to be linearly proportional to the gas concentration.

While the prototype uses a helium-neon laser, Kamm sees the system eventually packaged into a suitcase, by substitution of a tunable solid-state laser for the helium-neon unit.

The system has been demonstrated to the Federal Law Enforcement Administration for detection of the minute amounts of vapor given off by explosives and heroin.

The Advanced Projects Agency sees another use of the technique: detection of atmospheric absorption of infrared laser beams used for communication and navigation systems. ■■

Space antennas use quasar signals to predict earthquakes

Radio signals from outside our galaxy will soon be used to detect almost imperceptible movements in the Earth's crust that may eventually lead to the accurate prediction of earthquakes.

In a project dubbed ARIES, for Astronomical Radio Interferometric Earth Surveying, scientists from NASA's Jet Propulsion Laboratories in Pasadena, Calif., have taken instruments originally developed for space exploration and adapted them for use in an earth fault-monitoring system. By combining the experience they've gained from spacecraft navigation with current radio astronomy and geophysical research, JPL scientists hope to be able to measure movements as small as a few inches in the earth's crust.

According to Peter Mac Doran, the project director, the principal

instruments in the earthquake research project are a 30-ft. dish antenna located at JPL's Pasadena facility, and a 64-meter dish located in the Mojave Desert in Goldstone, Calif. According to Mac Doran, a sensitive radio receiver linked to the antenna listens to signals originating from quasars that are as much as one-billion light years away.

One of the space-communications stations of NASA's Deep Space Network in Goldstone will listen to the same signals, Mac Doran continues. Success of the project, he states, is linked to JPL's ability to measure the difference in the arrival times of identical quasar radio signals at the two dishes located about 125 miles apart.

Mac Doran is confident that the researchers will be able to make

these measurements without too much difficulty. In explaining why, he says that over the past three years, instruments originally developed for space navigation and communication have been developed that produce time measurements with a precision of 1/10 of a billionth of a second (1×10^{-10} sec).

In explaining how the system works, the director notes that S-band signals from a quasar are received and heterodyned with a signal from a local oscillator to produce an output signal in the 0-to-2-MHz range. This signal is digitally sampled at a rate of 4 Mb/s, and then "tagged" with a time code, and recorded on a magnetic tape recorder. Tapes from the two stations are then fed into a computer, where a cross correlation is performed to get the difference in the time of arrival

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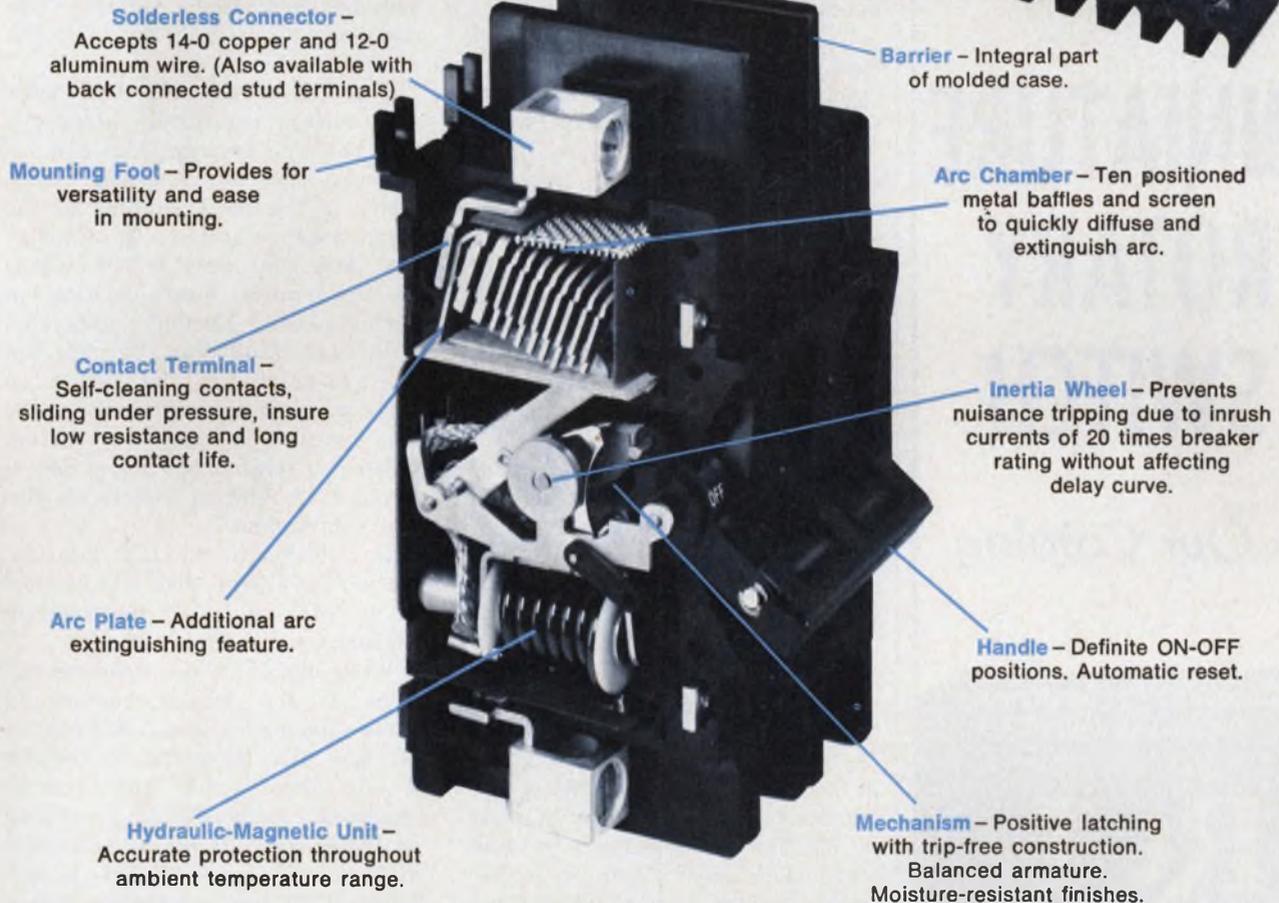
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INFORMATION RETRIEVAL NUMBER 29



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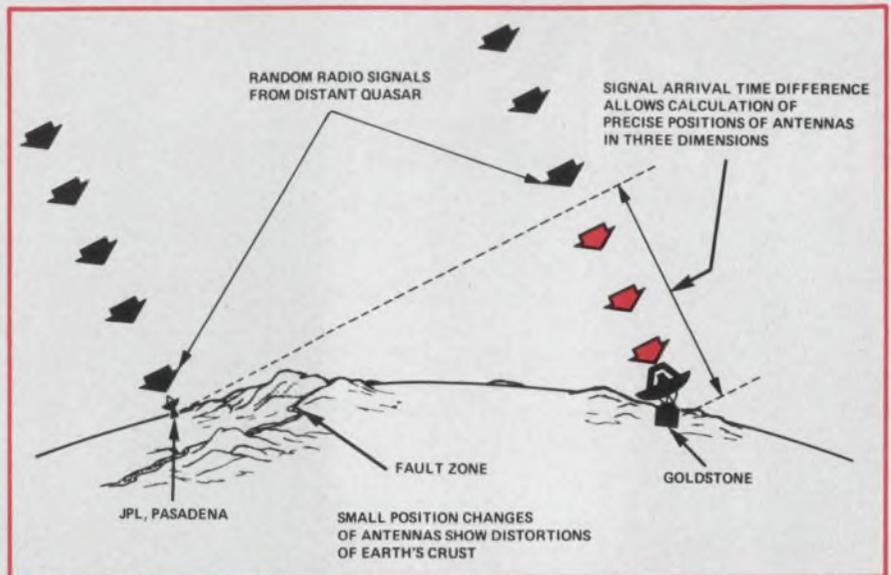
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Prediction of earthquakes is the goal of Project ARIES, which uses signals generated by quasars to detect movement in the earth's crust.

of the signal at each location.

By looking at five or more quasars, enough data are generated to determine accurately the three-dimensional separation of the two antennas, Mac Doran says.

To make sure that the clock and local oscillator signals at both locations are the same, a rubidium atomic clock is used as a reference.

Commenting on the vast difference in size between the two dish antennas, Mac Doran notes that the results of this system depend on the geometric mean of the receiver sensitivity and the collecting area of the antennas. Thus the big Goldstone dish, with its maser amplifier, compensates for the smaller JPL dish with its parametric amplifier.

An initial distance measurement between JPL and Goldstone will establish a reference baseline. Subsequent measurements will indicate any change in the distance between the two dishes that might be caused by shifts in the earth's crust.

Explaining how the motion of the earth's crust is detected, Mac Doran notes that any deformation in the earth between these two antennas will cause them to shift in location with respect to each other. This shift from the reference positions, set up at the beginning of the experiment, will cause radio signals from the quasar to arrive a fraction of a second before or after the time they should have normally arrived. This fraction-of-a-second difference

in time is measured, and the amount of movement is calculated.

With the high-precision equipment available, changes in the distance between the two antennas can be determined in three dimensions to an accuracy of a few inches or less.

The Goldstone and JPL sites were chosen specifically because a straight line between the two intersects the St. Andreas fault, which is the main channel of the earth-fracture-system in California. And that area is the source of the state's most devastating earthquakes. Motion associated with this fault can be detected right away, and can be used as an early warning against earthquakes. The reason is that the motion creates a strain that must be relieved by a mild or sometimes violent earthquake.

In addition to the JPL location, the portable 30-ft. dish will be used in at least six other locations in southern California.

With its ability to measure motions of the earth's surface in three dimensions, the ARIES project may also be useful to confirm a new theory that some earthquakes are preceded by a swelling of the surface of up to 3 ft. over an area of hundreds of square miles. This rock-expansion theory is based on the belief that stresses in the crust cause minute cracking of rock on a vast scale. This tends to lift the earth's surface. If the theory is confirmed, earthquake predictions would be better. ■■

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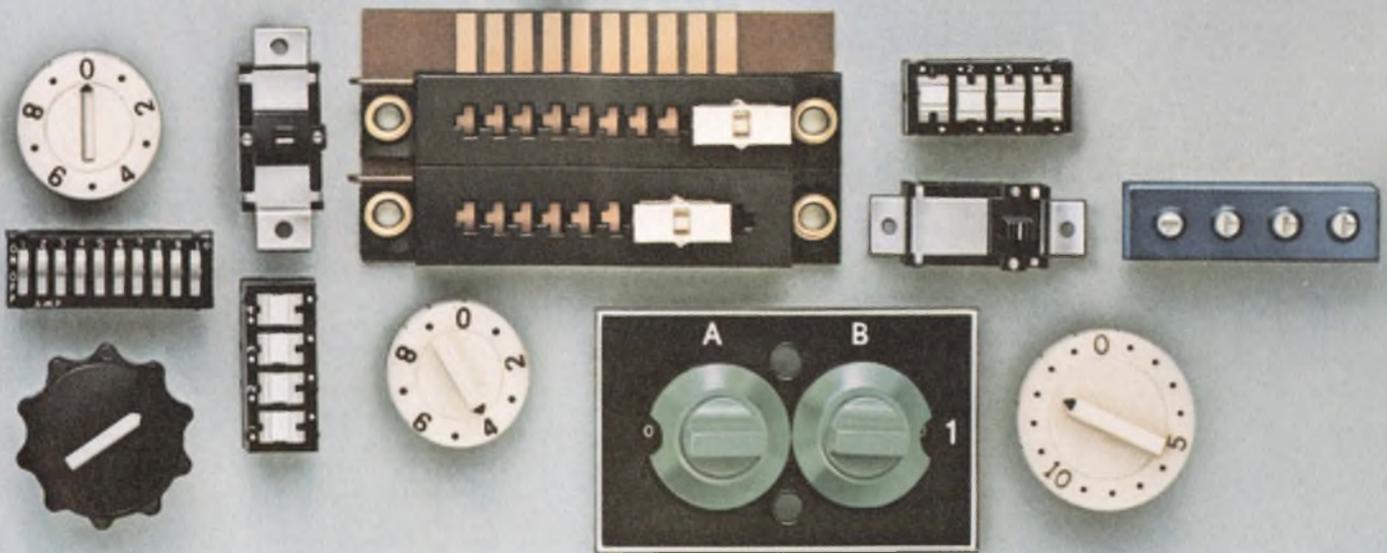
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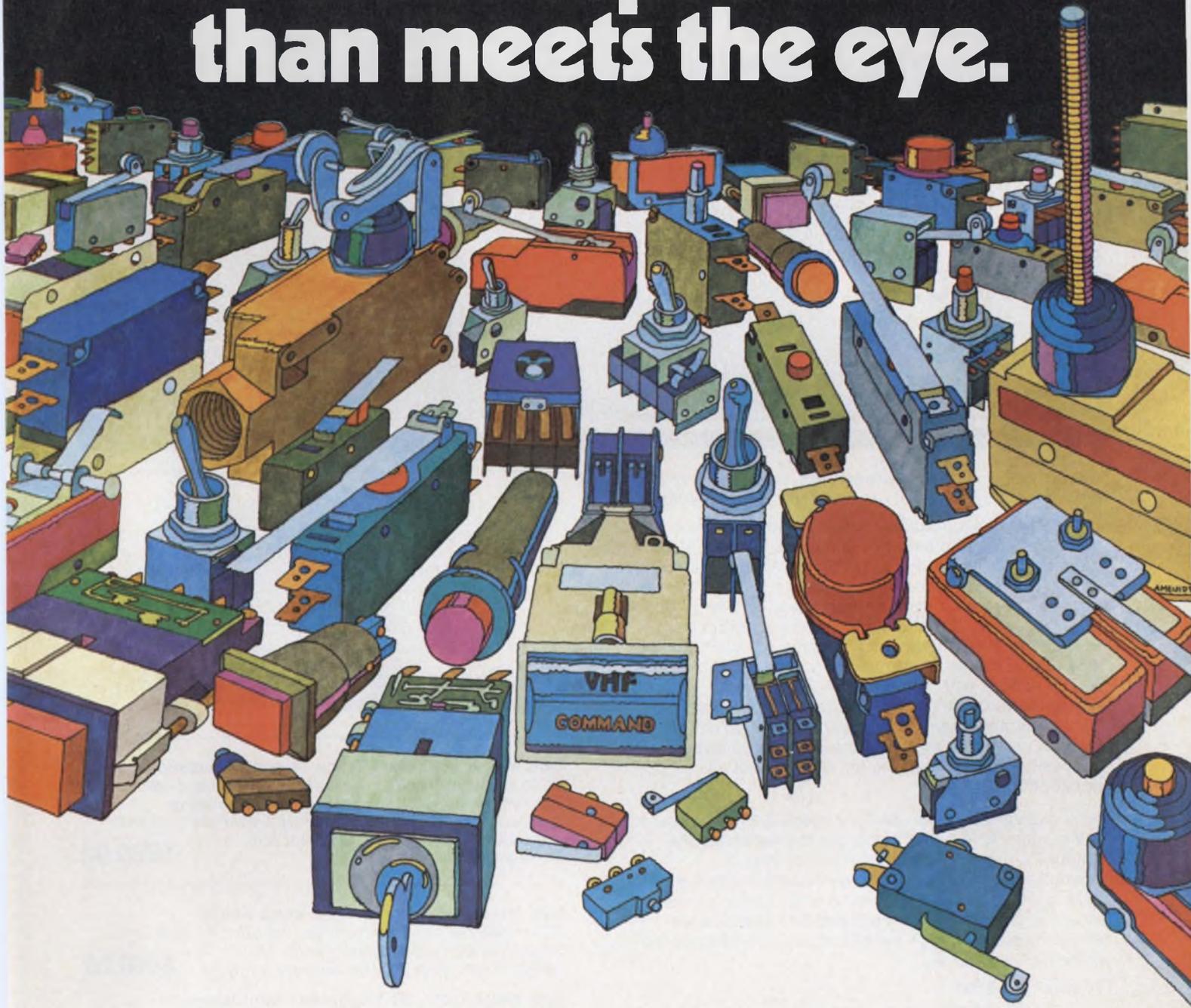
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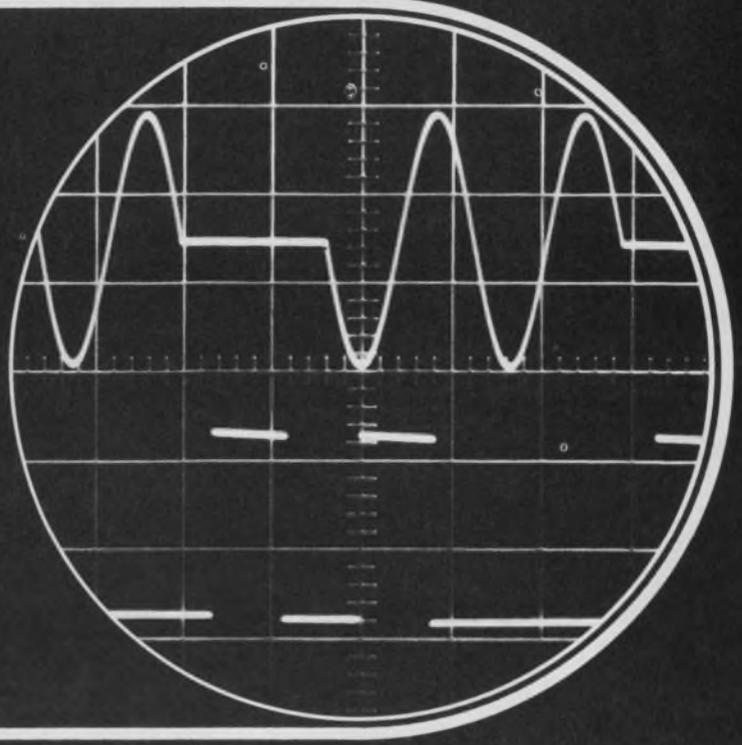
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INFORMATION RETRIEVAL NUMBER 37



washington report

U.S. Pension Reform Act: A start, anyway

President Ford's signing into law the Pension Reform Act was a milestone for some 30 million American employees but hardly the last shot for pension reformers, who won't be satisfied until pensions are portable from job to job and the pension plans are mandatory. Yet the bill made giant strides in stipulating acceptable vesting systems. It specified minimum funding and fiduciary standards, and it established a Pension Benefit Guarantee Corporation to insure pensions.

For the self-employed, contributions to the so-called Keogh type of private pension plan were raised to 15% of annual earnings or \$7500, whichever is less—up from the old \$2500 maximum. New in this category of plans is a provision that allows employees not covered by tax-qualified government or annuity pension plans to create their own retirement accounts. The annual limit is 10% or \$1500, whichever is less.

A possibility for portability is written into the new reform law. With a former employer's approval, an individual may transfer vested benefits, but he must do so within 60 days after leaving. Conceivably these funds could then be transferred to the new employer's pension plan, if he was agreeable. This and other provisions of the law are murky, and experts themselves aren't sure on many aspects. It will probably be months before the full impact of the bill is known.

Long-range study of microwave effects due

The increasing use of microwave-generating devices—and the flap over microwave ovens—has prompted the National Institutes of Health to seek sources to investigate the effects of long-term, low-level pulsed and modulated microwave radiation. The focus is on the 1-to-10-GHz frequency range and possible impact on the functioning of the nervous system. Officials say the objective is to learn more about biological effects, which appear to be more pronounced with length of exposure.

U.S. calls metric conversion inevitable

Although it was a backdoor operation, the U.S. does finally have a metric policy of sorts that indicates a recognition of the inevitable.

In an amendment to the Elementary and Secondary Education Act, recently passed by Congress, is an authorization for \$50-million to be used at the discretion of the U.S. Commissioner of Education for metric education through fiscal 1978. The amendment states that increased use of the metric system is inevitable and will become the dominant system of

weights and measures in the nation.

An added push for Congress to pass a metric bill that would coordinate conversion is coming from an agreement being implemented by the European Common Market. By 1978, U.S. exporters will be required to indicate dimensions in metric units.

Expansion for automatic vehicle location

The Federal Communications Commission, in recognition of the dramatic developments in automatic vehicle location, is amending its rules to accommodate more users by opening new frequency bands. In general, narrowband base mobile data links in automatic vehicle monitoring (AVM) systems are being developed on uhf frequencies in the 450-MHz range. Various proximity sensing systems also use frequencies at 150 MHz or the 27-MHz band, but there has been a problem in meeting channel requirements because of the narrowband spectrum available.

To alleviate this condition, the FCC is reallocating 902 to 912 MHz and 918 to 928 MHz for wideband AVM applications employing pulse-ranging, multilateration techniques. This is where vehicle positions are determined by the difference in time of arrival, or phase difference, of signals received at fixed sensors or at the vehicle. Such systems are expected to handle up to 10,000 vehicles and be suitable for large or multiple users.

The FCC reports that AVM techniques are now being used in Chicago's bus transportation system and in police operations in St. Louis.

Capital Capsules: The Air Force is planning to evaluate reliability and associated failure mechanisms for complementary MOS transistors using silicon on sapphire technology. . . . The Air Force intends to conduct a comprehensive engineering study of electronic counter-counter-measures (ECCM) techniques applicable to digital tropospheric scatter radio communications systems. Included will be scatter and various means of line-of-sight jamming. . . . The Dept. of Defense is planning to issue a six-month contract for technical assistance to the National Military Command System Support Center for analysis, design, development and maintenance of selected data files and associated hardware. Bidders must have extensive knowledge and experience with IBM 360/67, HIS 6000, and CALCOMP plotter. . . . The Navy is seeking several-source production of OE-82B/WSC-1(V) shipboard antenna systems used for satellite tracking. The system uses two antennas and diplexers to give simultaneous transmission and reception, regardless of the ship's bearing in relation to the satellite. . . . The Air Force is looking for companies to develop a sophisticated damage-assessment system. Involved are theoretical and analytical investigation of remote-damage assessment of airborne vehicles by measurement of trajectory and by multispectrum signature analysis. . . . The U.S. Bureau of Mines has taken delivery on an electronic gas analyzer, "blue box," that will be used to detect dangerous concentrations of explosive methane gas or unusual accumulations of carbon dioxide in coal mines. The bureau is building a capability to determine electronically whether a mine's ventilation system is operating correctly, whether there is a fire or if a known fire has been extinguished. . . . The Atomic Energy Commission has issued a Request for Proposal for the design and construction of one or more Demonstration Centrifuge Enrichment Facilities, to be in operation by the late 1970s. The AEC concludes that small demonstration plants on private sites is Step 1 in luring private firms into the business.

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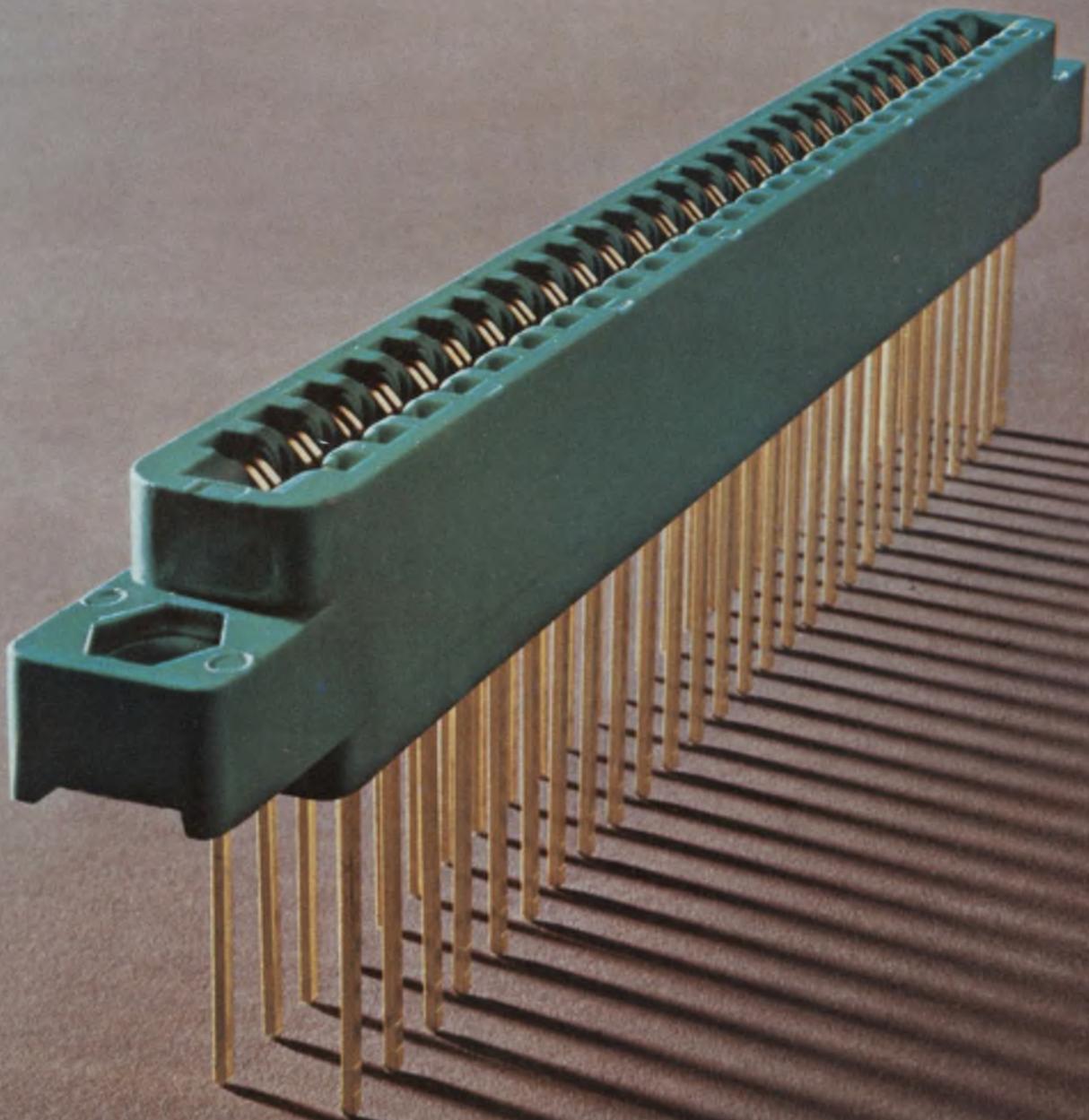
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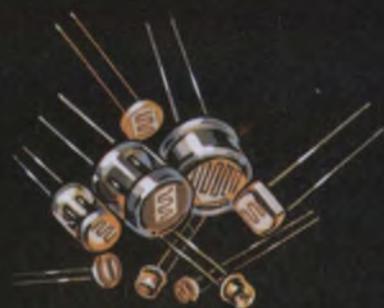
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editorial

It's a fine first baby. Father's moderately pleased.

Many of you have been kind enough to let us know how useful you found Electronic Design's GOLD BOOK. We really appreciate those kind words, for directory editors rarely hear from their readers.

Frankly, we really worked at the GOLD BOOK. I use the word "we" loosely here. What I really mean is George Weingarten, the GOLD BOOK's editor. For a while, George spent so much time at the office that I'm sure his wife, on his rare visits home, had to re-introduce him to his kids.

The first edition was a mammoth undertaking. It has information on 7500 manufacturers, with sales outlets listed for 3100 of them. For some 3000 products, it lists the manufacturers with complete address, zip and phone number—not just the city and state. And for each product, it shows which manufacturers have provided verifying literature. It lists 5700 distributors and 4500 trade names—in addition to 2800 pages of catalog data. In every way we could conceive, it was designed to make it easy for a fellow to locate the sources of the products he needs.

Now that the first edition has been distributed, you'd think that Weingarten, his right arm, Editorial Assistant Edith Merker, and I would be overjoyed. We are, in fact, rather pleased—but not content. We've already found several errors and we're sure there are more. We didn't expect it to be perfect, though we tried to make it so. We already have some ideas for making it better next year. But we're annoyed with ourselves for not having thought of these for the first edition.

What really irks us, though, is that we're sure there are ideas that will come to us after we have put the second edition to bed, and plenty of other ideas that may never occur to us.

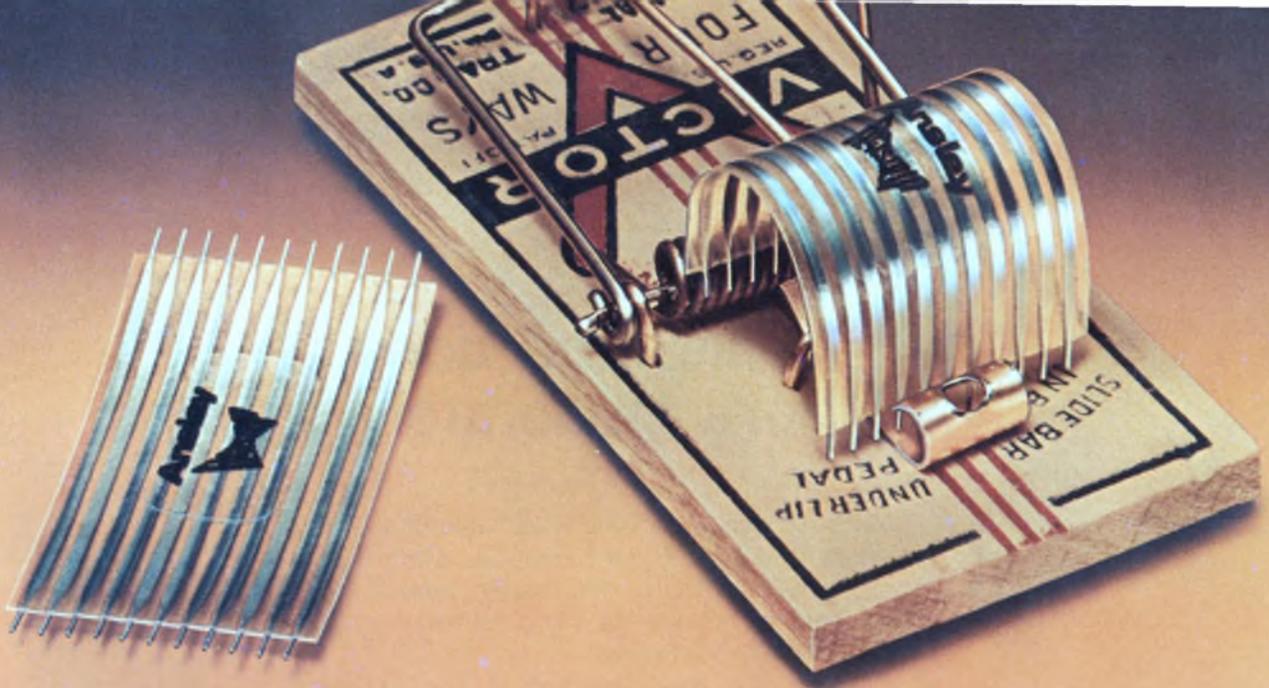
We would genuinely appreciate your help. As you use the GOLD BOOK, if you see some ways we can make future editions more useful, please drop us a note or use the card following page 160A. We'll probably never be completely satisfied. But we hope each year's edition will be better than the earlier one.

Thank you.



A handwritten signature in dark ink, which appears to read "George Rostky". The signature is written in a cursive style and is positioned above the printed name.

GEORGE ROSTKY
Editor-in-Chief



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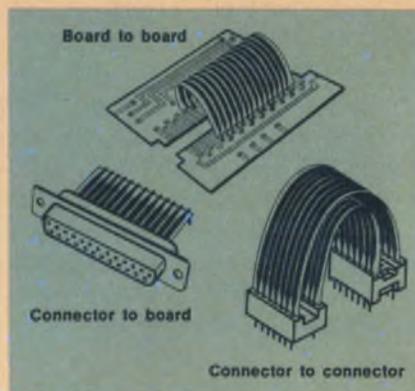
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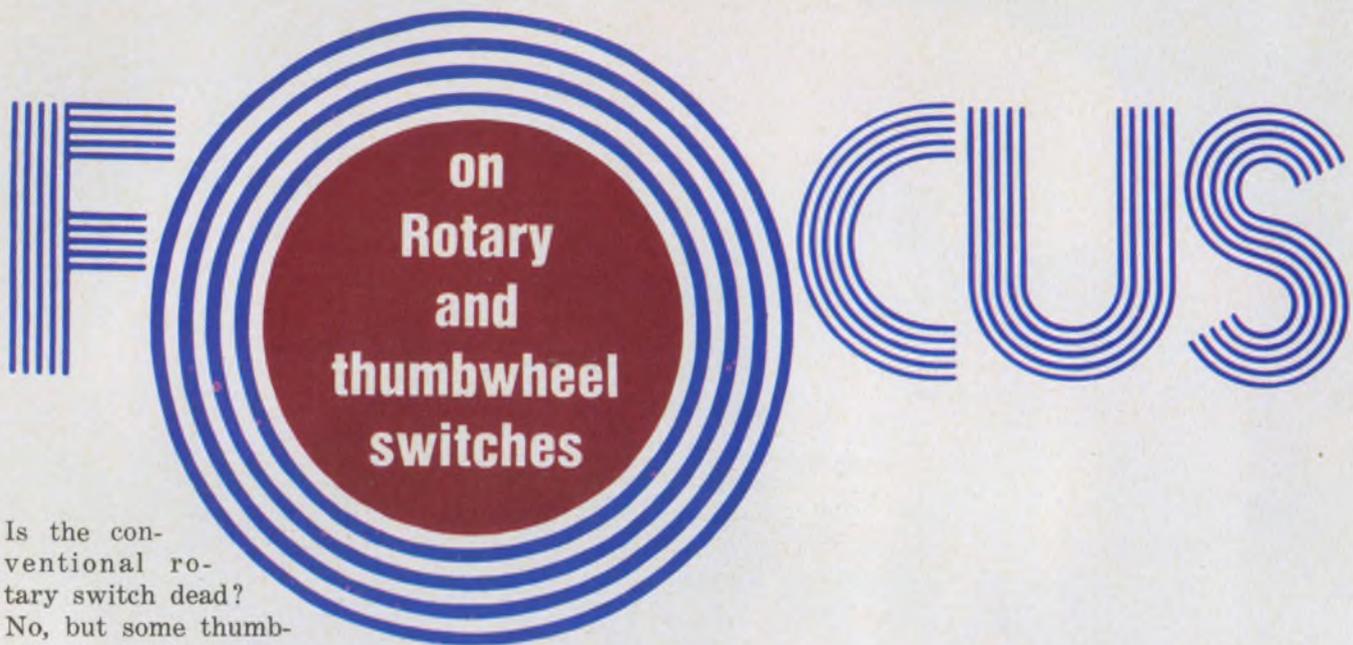
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INFORMATION RETRIEVAL NUMBER 45



FOCUS

on
Rotary
and
thumbwheel
switches

Is the conventional rotary switch dead?

No, but some thumbwheel-switch manufacturers wish it were.

The rotary remains the workhorse of selector switches, while the thumbwheel has moved into more sophisticated applications.

And all types of selector switches—from conventional rotaries to the latest thumbwheel and level and pushbutton versions—are alive and well.

Straight thumbwheel switches are replacing the ordinary rotary in many uses. Meanwhile lever and pushbutton “thumbwheel styles” are making inroads into the market of the “straight thumbwheel.” With an over-all expanding electronics market, sales of all types are increasing.

The result: The designer has a wide, very competitive field from which to choose a selector switch. But he also has more ways to get stung. Manufacturers' catalogs are the first step into the hornets' nest.

Catalogs can steer you wrong

Catalog listings for selector switches can be very misleading. In attempts to simplify spec listings, switch manufacturers compare switch designs in an arbitrary way. The criteria for their comparisons rarely match the needs of actual applications. If 49-m Ω contact resistance means that the switch is good, can 50 m Ω mean that the contact has failed? Some tests call for exactly this kind of end-of-life determination.

Most reputable switch manufacturers have compiled tons of data on their products. But the data are based upon accelerated life tests and arbitrary end-of-life criteria. They are difficult to relate accurately to the real world of applications. Fearing that their data will be misinter-

preted, manufacturers often go to the extreme of providing too little information in catalogs.

It is true that the many possible switch requirements, coupled with the many design solutions, can result in almost infinite permutations of ratings. The effects of load, duty cycle, environment, contact material, insulators and other factors on failure criteria are complicated. But most catalog listings show load and life ratings for so limited a range of conditions, that the data seldom are sufficient for a rational choice.

How do you determine which switch will perform reliably in your application?

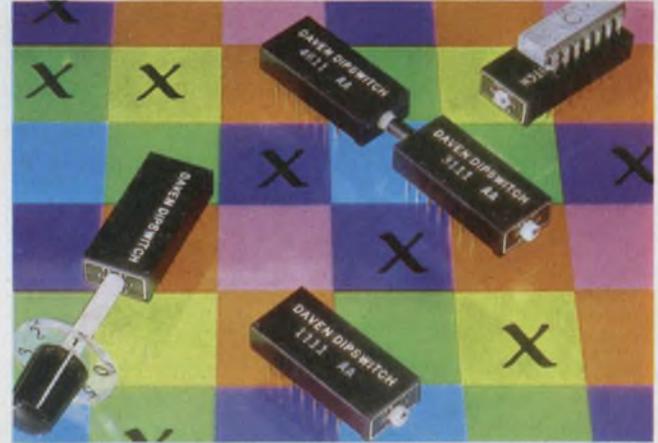
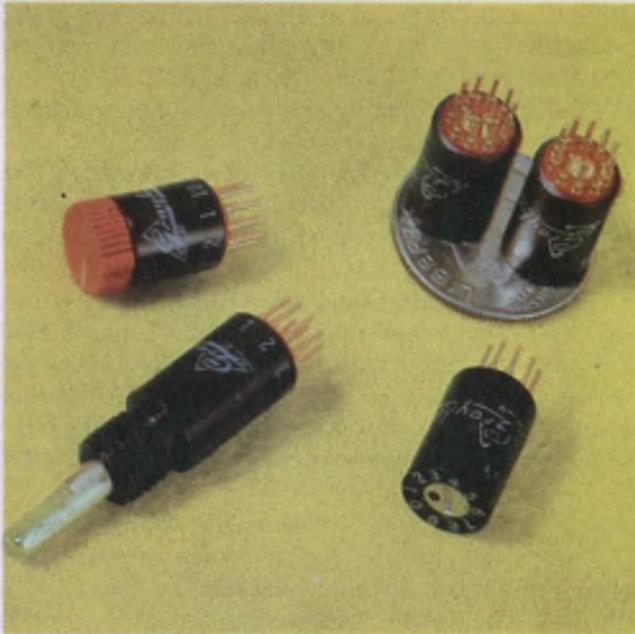
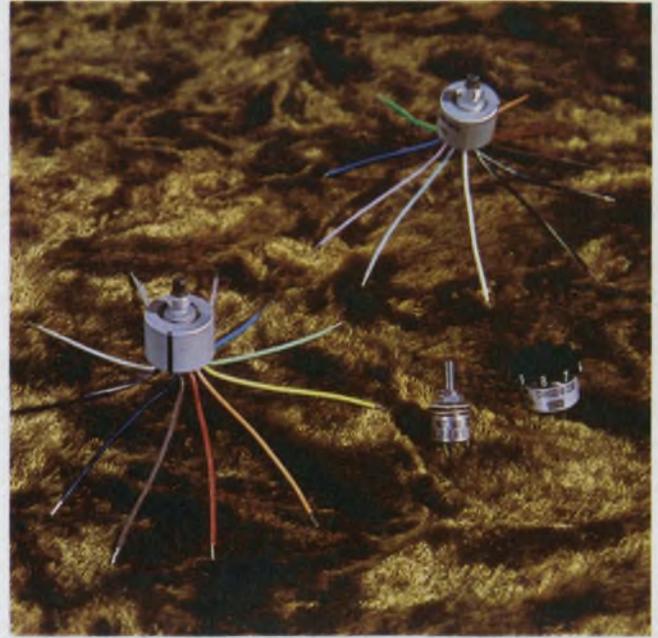
Decisions, decisions

First, decide on the style of selector—conventional rotary, thumbwheel or one of the “thumbwheel” variants. Rotary switches, in general, offer several advantages over nonrotary. They can control many circuits simultaneously—and in a tremendous variety of sequences. And most rotaries, except perhaps the lever, are more difficult to actuate accidentally than toggle switches. Because of this, equipment designers often use a two-position rotary instead of a toggle switch.

If the design calls for the handling of extremes in voltage, current and environment, the conventional rotary styles will fill the bill. At signal-power levels in the more complex circuit combinations—like encoding and decoding—most designers use the thumbwheel variety. For in-between applications—range and function selectors, for example—both rotaries and thumbwheels are used.

Conventional rotaries usually occupy more panel space than thumbwheel types to do the same job. And when rapid positioning is desir-

Morris Grossman
Associate Editor



Miniature rotary switches for PC-board mounting include (upper left and clockwise) AMP 642-1 series units, the Minelco SW 37 and 67, the Edison, Daven Dip-switches and the Grayhill Series 75. Other rotary-switch manufacturers are turning out similar types in an almost endless flow.

able, variations on the thumbwheel style that use lever actuators are available. Operation of the switch with gloves might dictate a pushbutton or level variety. Finally, "human engineering" and esthetics—where visual, auditory and tactile considerations are of prime concern—might be the ultimate arbiter in the selection of a switch style.

After you've decided on the switch style and made an initial selection from the meager data in most catalogs, draw up the following list of requirements and submit them to several manufacturers:

- Application—type of equipment.
- Switch drive—manual or motor driven, and duty cycle, drive torque and stop limits.

- Load—voltage, current, frequency, power factor.

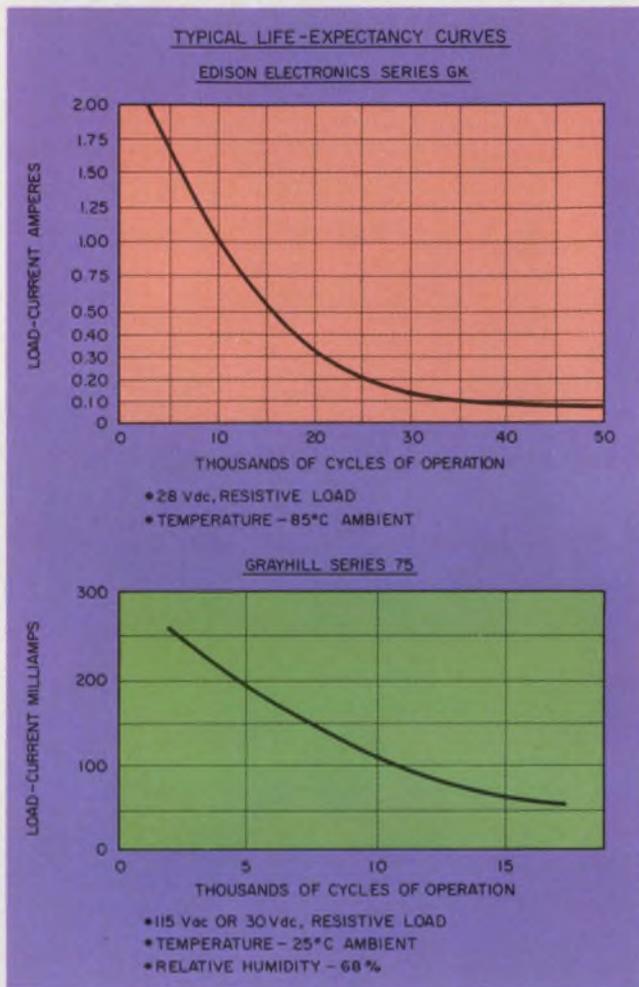
- Environment—temperature, altitude, relative humidity, shock and vibration, corrosive atmosphere, etc.

- Expected life—in number of cycles per agreed failure criteria.

- Failure criteria—contact resistance, insulation resistance, dielectric strength, etc.

A reputable manufacturer can then estimate whether a given switch selection is the right one for your application, or he may recommend a better one.

But before you can talk intelligently to the switch manufacturer, you must learn to specify in his language.



The "useful" life of a switch in your application depends upon the demands the circuit makes upon the switch's load-carrying ability, its contact resistance, insulation resistance, dielectric strength and detent mechanism.

A switch's life is hard to measure

What may be totally unacceptable in one instance may be perfectly suitable in another. For example, a contact resistance of 0.1 Ω might be unusable in the range switch of an ammeter. However, even 1 Ω would make no difference in most logic-circuit selector controls.

And don't make the common mistake of selecting a switch based only upon its performance

when new. Most switch characteristics deteriorate with use and time. Some switches die gracefully. Others drop dead without warning, as when a spring breaks in the indexing mechanism. More important than initial characteristics may be the performance of a switch after several years of use.

Let's examine some of the criteria that manufacturers use to determine "end of life."

Contact resistance, under correct electrical loading changes gradually. Low resistance is the prime requirement of a switch contact. Typical values for rotary switches lie in the milliohms range. For example, Grayhill's Series-75 rotary switch is specified at a maximum of 10-m Ω initial contact resistance and an end-of-life maximum of 50 m Ω . Thumbwheel switches usually have higher initial and final values. Their field of applications—mainly in logic circuits—doesn't require extremely low contact resistance. Typical is EECO's 8000 series thumbwheels. They have an initial 100 m Ω maximum and a specified end value of 160-m Ω maximum.

Contact resistance and the other life criteria are measured as the switch is cycled. A cycle is defined by most manufacturers as one complete rotation through all contact positions and a return through all switch positions. But watch out. Some vendors may wish to impress you with large numbers and use individual operations, or index steps.

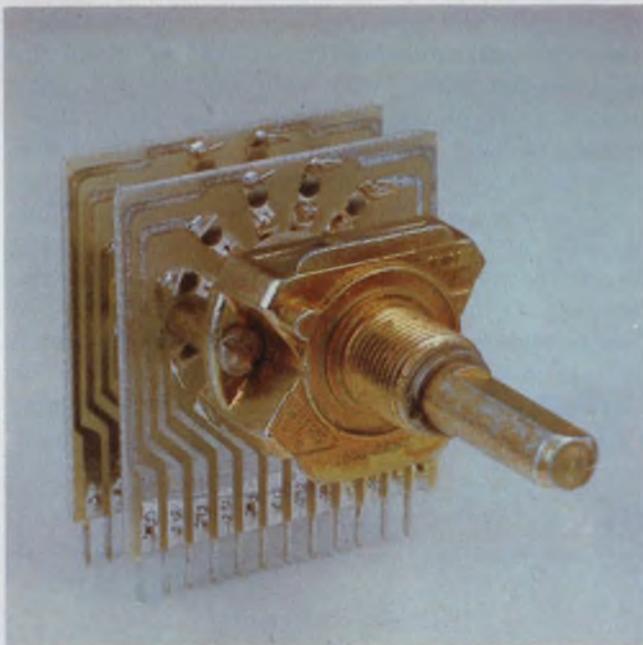
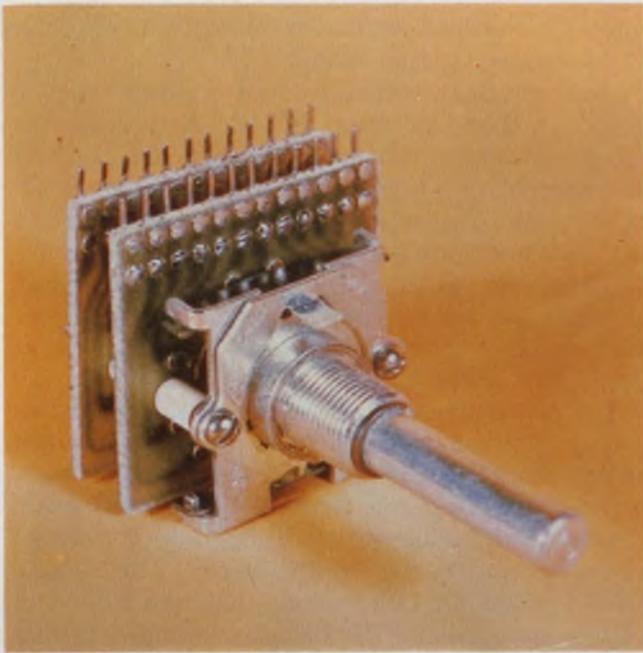
Thus, on Edison Electronics' Series GK switch, the life curve (see plots) correctly shows that 200 mA can be switched for about 25,000 cycles before the contact resistance, or one of the other failure criteria, terminates the test. This 12-position switch then moves through $2 \times 11 \times 25,000 = 550,000$ index steps.

A noncatastrophic reason for failure, arbitrarily established by the manufacturer or perhaps a MIL-spec, usually signals the "end" of a life test. The switch seldom disintegrates when it reaches this "failure" criterion. For a great many applications, it could still be quite satisfactory—its life continuing for many more cycles until some mechanical part breaks.

So-called "life-expectancy" data apply only to the narrow conditions of the test. The values obtained may be far too optimistic or pessimistic, to determine the actual useful life.

Short table of MIL-Spec end-of-life criteria

| | Contact resistance m Ω | Insulation resistance—M Ω | | Dielectric strength V _{rms} |
|------------|----------------------------------|----------------------------------|------------------|---|
| | | dry | 90-98% RH | |
| MIL-S-3786 | 20 | 1000 (plastics) | 10 (plastics) | 1000 (1-mA leakage max) |
| MIL-S-6807 | 20 | no spec | 3 (after drying) | 600 (10-mA leakage max) |
| MIL-S-8805 | 40 | 1000 | 10 (plastics) | 1000 + 2 \times WV (1 mA leakage max) |



The popularity of printed-circuit wiring has prompted many conventional rotary-switch makers to convert some designs to PC styles. Centralab's Series 160 (top) and Oak's 12-position rectangular units have all terminations along one edge to facilitate PC soldering, and Stackpole (bottom) follows suit with a miniature, enclosed unit, the Series 80.

Thus the life-expectancy curve of the Grayhill Series 75 switch is only for 115-V ac or 30-V dc resistive loads. And the ambient conditions for the manufacturer's test were sea level, 25 C and 68% relative humidity. For other temperatures, altitudes, humidities and loads, most manufacturers advise: "Ask us."

Insulation a factor in failure

Reduction of insulation resistance or dielectric strength—which is frequently catastrophic—also signals what manufacturers and the military call end of life.

Insulation resistance is measured between any two insulated metal parts that may become part of a circuit. For new switches, typical values are in thousands of megohms. The resistance of an insulator, especially with surface contaminants, tends to be nonlinear; thus measurements are made at specific high dc potentials—usually 100 or 500 V for signal-level switches.

As an example, RCL's Series FC and RC commercial switches have an initial insulation resistance spec of 50,000 M Ω . When the resistance drops to a minimum of 10,000 M Ω , the switch, according to the specs, is at its end of life. But 10,000 M Ω may be more than adequate for low-impedance, low-voltage circuits, and many more cycles may be tolerated before the switch is dead.

The insulation resistance between contacts drops with use, mainly because of the buildup of surface contaminants from the environment and from metallic particles that rub off the contacts onto the insulation. Moisture also affects insulation resistance. A test condition commonly specified is the measurement of insulation resistance in high humidity (90 to 98% relative humidity). In this atmosphere there is usually condensation on the insulation. Some insulators absorb moisture, and reduced insulation resistance results.

Though surface contaminants can often be removed to restore the insulator's resistance, life-expectancy tests usually don't permit this procedure. Thus, again, the accelerated tests tend to provide more pessimistic data than actual use might justify.

In addition to reduced resistance, insulators also suffer from dielectric breakdown. But in an actual voltage breakdown the odds are high that the switch would be damaged. Thus so-called "breakdown" tests are only checks for "excessive" leakage current at some high ac voltage. A new selector switch might withstand 1000 V rms, with only microamperes of leakage current. But deterioration of the dielectric material with age and temperature can result in a typical end-of-life maximum leakage of over 1 mA (see MIL life table).

The selection of insulation material for commercial applications is often first based upon price and then personal preference. In military applications the MIL spec may dictate the insulator material (see selection chart). There is a broad overlap in characteristics among the choices. Phenolic insulation, because of its low cost, has been very popular for industrial use. However, diallyl phthalate, which as the chart shows offers superior qualities, is gaining in use.

Torque specs are inadequate

Of the life criteria, rotational torque, or actuation force, is probably the more inadequately treated. Index mechanisms deteriorate with use and can end the life of a switch before anything else fails, though frequently mechanical life is longer than electrical life under full load.

A CTS 211 index assembly is rated for 250,000 cycles through 12 positions and return. This is perhaps five times the electrical life of some of the contact assemblies that could be used with it.

The mere listing of maximum and minimum torque is not enough to establish end of life. Two offsetting factors operate to change the initial torque values during the life of the switch. Wear and galling of the switch's mating surfaces tend to increase the force needed to turn the switch, and fatigue and relaxation of spring members lowers the forces.

Thus end-of-life criteria should call for a maximum allowed percentage change from the initial value. A typical spec might read: "The

rotational torque shall not change more than 50% from its initial value."

Most catalogs provide very skimpy torque specs, if they list them at all. And then the specs are only of the initial-value type. Here is an example of an inadequate torque spec: "20 to 80 oz-in., depending on the number of poles per deck and number of decks." Here's another: "As required within a range of 5 to 25 oz-in." You are not told what torque is required per deck or pole; what the tolerances are; or what the failure criteria are.

The construction of detent mechanisms varies widely between manufacturers. Even different switch lines of the same manufacturers have individualized designs. Some low-cost units use a circular, flat spring with a dimple, or V, formed at one point on its circumference. The dimple rides against an index plate that has hills and valleys stamped into it. Designs intended to last longer use a pair of steel balls or rollers instead of a dimple; this provides a more positive indexing action and more predictable torque. When the balls are combined with coil springs and with precise index plates for the balls to ride against, predictable and adjustable torques, long life and superior "feel" can be obtained.

Of course, you pay for these features. Most manufacturers prefer that you not specify torque, because it is difficult to control. Centralab says that its standard torque tolerance on rotary switches is 20% or ± 10 oz-in., which ever is greater. But 10% or ± 5 oz-in. can be attained on some of Centralab's index designs at

Rotary switch insulator selection chart

| Stator Insulation | MIL Spec | Usable temperature limit (°C) | Relative cost | Minimum insulation resistance M Ω | Important characteristics |
|-------------------|------------------------|-------------------------------|---------------|--|---|
| Phenolic | MIL-P-3115, type PBE-P | 100 | 1 | 1,000 | Most economical, most commonly used. Though nonnutrient to fungi, varnish impregnation per MIL-V-173 is available. Meets most military and commercial application requirements. |
| Diallyl Phthalate | MIL-M-14F | 100 | 6 | 10,000 | Excellent structural and dielectric strength. Low moisture absorption rate. Not affected by most solvents and flux cleaners. |
| Epoxy-Glass | MIL-P-18177, type GEB | 125 | 4.5 | 10,000 | Economical, excellent structural and dielectric strength. Lower moisture absorption rate than phenolic. |
| Silicone Glass | MIL-P-997C, type GSG | 125 | 4.5 | 10,000 | Good RF characteristics. Lower structural strength than epoxy glass, and slightly higher in cost. |
| Mycalex | MIL-I-10A, Grade L-411 | 150 | 10.5 | 10,000 | Good structural and dielectric strength, lowest moisture absorption. |
| Kel-F | MIL-M-55028-A Grade 3 | 125 | 9 | 20,000 | Excellent structural and dielectric strength. Low moisture absorption. |
| Ceramic | MIL-I-10A, Grade L-422 | 150 | 3 | 10,000 | Very low dielectric loss and moisture absorption. |

Courtesy of Ledex Inc.

a "considerable" cost increase.

Conventional rotaries can be specified with a large variety of shafts: round, single-flat, double-flat, knurled-slotted, insulated, etc. Of course, the dimensions and location of flats, if any, must be carefully specified by the user. Though a switch is an electrical component, the physical dimensions and mechanical details in catalogs are usually more adequately presented than are the electrical details.

Watch the load

Though most switch specs distinguish between current-carrying and current-switching capabilities, some don't. Ask if the information is not clear. An example of a clearly rated switch is Ledex's Series 200 rotary. The switches are rated to make or break 0.5 A at 28 V dc or 0.25 A at 110 V ac, resistive, for silver-plated brass contacts. The contacts can carry 5 A.

The carrying capacity is higher than the switching capacity, because an electrical arc is drawn when a circuit is broken. And, of course, inductance in the circuit increases the arcing problem when a circuit is opened. On the other hand, a large capacitor with low series resistance can weld the contacts of a switch when a circuit is closed.

Often contacts are specified for handling inductive circuits, but rarely is an allowed value of inductance stated. A common standard, derived from MIL S-3786, is a value of 2.8 henries with an L/R ratio of 0.026.

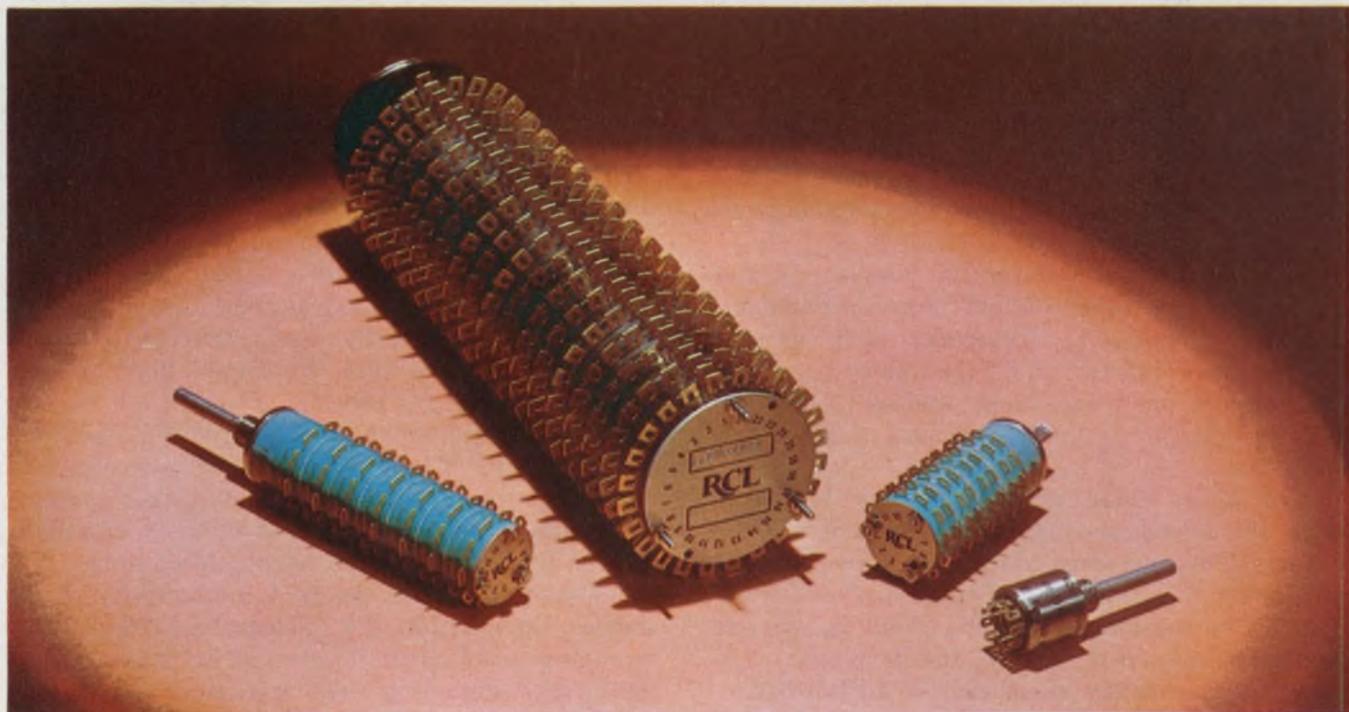


Conventional rotaries are the workhorses of the selector-switch world. These Centralab Series 500s are built with three different insulator materials. Many manufacturers allow a wide choice of insulator material for standard switch lines.

Resistive-load current capacities are generally five to 10 times larger than inductive. And resistive ac ratings are higher than dc, because an ac arc will extinguish itself when the current goes through its instantaneous zero, while a dc arc tends to maintain itself. The higher the frequency, the sooner the arc is extinguished, with less erosion of the contact.

A fast-acting index mechanism, which tends to snap the contacts open rapidly, can reduce arc damage. Conversely, a switch that can be "teased" is undesirable. In addition to arcing, teasing can cause improper, even dangerous circuit operation.

Reduced barometric pressure at high altitudes may worsen the arcing problems. Air at low pres-



Completely enclosed rotary switches are protected against most environmental hazards. According to RCL,

enclosed rotaries are price-competitive with the conventional open-frame types.



Modified "thumbwheel" switches use pushbuttons or a lever to move the contacts. The lever-actuated switch from Cherry (middle left) rotates the switch through all 10 contacts with only 60 degrees of motion. The push-button styles from Alco (top left), Janco (middle), Durant (bottom left) and Oak (right) move one step at a time.

sure ionizes at low voltage and maintains the arc longer. Thus current and voltage ratings must be derated at high altitudes.

The amount of lubricant initially put into rotary and thumbwheel switches during manufacturing is generally sufficient for the life of the switch under normal room temperatures. But continued operation at elevated temperatures results in gradual lubricant loss, with damage to the mechanical parts.

High temperature reduces the viscosity of the lubricant in the bearings, and index mechanism. Standard commercial lubricants will usually operate from -45 to 125 C; for military service, the lubricant must withstand -65 to 125 C. For some special applications, lubricants to 175 C are available.

Used above their rated maximum temperatures, the lubricants may liquefy and flow onto contacts. This increases contact resistance, causes intermittent connections and traps dust and wear particles on the insulation. The result is lowered insulation leakage resistance. It's clear that high temperatures usually mean shorter switch life.

Accelerated life tests may not be much help here either. The tests are run at about 10 cycles

per sec, but manual operation of a switch rarely approaches this. The test speed tends to prevent the lubricant from redistributing itself over bearing surfaces as the switch is cycled. Heat from friction and arcing accumulates. It's obvious that the switch runs hotter than it would under manual use.

Dry circuits need gold

For most applications, the self-wiping action of conventional rotary switches allows use of economical silver-plated brass for the contacts and wiper blades. Silver-plated brass with 0.0001-in. minimum plating will last for about 10,000 cycles, but a solid silver alloy can run over 200,000 cycles. Thumbwheel types, however, because of their PC construction, lighter contact pressures and generally smaller current ratings, frequently have a thin, hard-gold layer plated over the copper PC base.

Typical is the 0.0002-in.-thick gold-plated contacts on Inter-Market Type-I thumbwheels. The rotor pickups are a silver-gold alloy over springy beryllium-copper. To help control wear, the alloy pickup is softer than the hard-gold contact.

Though silver has the best conductivity of all metals, the addition of a gold flash of 0.00003 in., minimum over silver is advised by many manufacturers where long shelf storage or infrequent use of the switch is expected. This is to avoid oxide buildup on the silver. For switches that



Thumbwheel switches like these Digitran units stack readily and can squeeze a lot of control information

will see steady use, a 0.0005-in. hard gold alloy, rolled onto the contacts, will last for 50,000 to 100,000 cycles; for the balance of the switch's life the contacts will operate on silver-to-silver surfaces.

Gold plating is almost a must for so-called "dry" circuits. But what is a dry circuit? Everybody talks about it; nobody defines it.

One pragmatic, but tongue-in-cheek definition is: "When the switch doesn't work well, it's a dry circuit." This definition is hardly useful for specification purposes, but it's about the best that you can get out of most catalogs.

Other definitions range from "relatively small current in a relatively low-voltage circuit" to an attempt at numbers—"a maximum of 1 mA at 1 mV," or perhaps "generally below 0.5 A at 28 V dc." None of these is very authoritative, of course.

This is an area sorely in need of clarification. The National Association of Relay Manufacturers—in "Engineers' Relay Handbook," published by the Hayden Book Co.—defines a dry contact as one that "neither makes nor breaks current." The book states that the widespread use of "dry" is "erroneous" for circuits or contacts that handle low-level voltages and current. On p. 92 the book discusses "softening voltage" of contact materials, and it ends the discussion with this statement: "Circuits in which there is no possibility of applying to the contacts either a steady-state or a transient voltage as great as the softening voltage are known as dry, low-level, circuits."

In electrical contacts the metal of the mating surfaces meets only at microscopic high points, which concentrate any current flow into high current densities. When the voltage across the contacts is above a certain value—the "softening voltage"—heat is generated faster than it dissipates, and the contact material softens. At the softening voltage, the area in contact tends to increase, molecular films are vaporized, contact resistance drops, and the voltage drop tends

into a small space. And spacers between switch units blend a stacked assembly into a unified whole.

to remain constant as current is increased.

Perhaps a better definition for "dry" could be based on the softening voltage of gold. Gold, which is recommended most often for dry-circuit contacts, has the lowest softening voltage of commonly used contact metals. Its value of 80 mV, with a small safety factor—to, say 50 mV—could then be the upper limit for the open-circuit voltage in a dry-circuit. Since the voltage across the contacts must drop below the open-circuit voltage when the contacts close, no current limits need be included in the definition.

MIL specs are, of course, a valuable guide to contact specifications, and many catalogs refer to them.

Know the MIL specs

MIL S-3786 is the primary standard for low-power rotary switches that carry less than 2 A. And MIL S-6807, 8805, 15473, 21604, 22710, among others, are applicable to other switches.

Many switches for severe industrial applications can be completely specified by reference to a military qualified-parts list (QPL) number. However, a QPL part will command a higher price than a commercial version. Though a commercial switch may be built to MIL designs, it doesn't come with the full testing, documentation and inspection certification that a full MIL-spec switch does.

Some manufacturers' catalogs reveal which types are MIL styles. Often you find out that commercial versions are available only by asking.

Conventional rotaries are the workhorses

In spite of many advantages claimed for thumbwheel switches, the conventional rotaries still fill most selector-switch applications. They handle very high voltages and current. An example is Ross Engineering's line of 10-to-30-kV, 25-to-50-A rotary switches. And at the low-power end, the Dipswitch line of Edison Electronics,

the PC-board switches of AMP (Bulletin 642-1) or Minelco's SW line are straight rotary switches for circuit-board use. Some miniatures use small conventional knobs, some built-in serrated wheels, and some versions need a screwdriver to turn them. Many other companies also have miniature switch lines; EECO's Stripswitch and Beckman's Model 374 are examples.

Conventional rotaries come in three basic contact types. All wipe as they make and break contact. This tends to keep the mating surfaces clean and to expose fresh surface for low resistance contact.

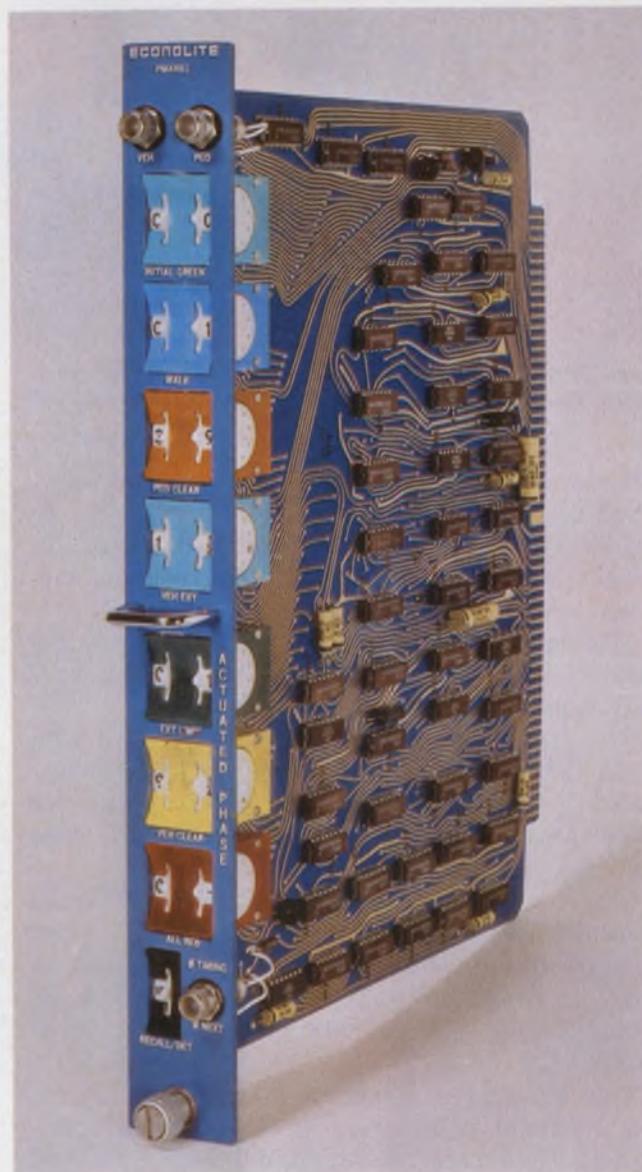
The most popular contact uses a rotary blade that wipes between stationary contact clips—examples include Oak's JKN series, Centralab's many SA types, and the CTS 223 selectors. In better types the rotary blade wipes between a pair of clips. The two independent mating surfaces provide far more reliable performance than a single-clip contact does.

The blade and clip-contact design comes in shorting and nonshorting styles. Shorting designs have a wide rotor-blade pickup member that can bridge the gap between adjacent clips and short them together when the switch is moved from one contact clip to the next (make-before-break). This is desirable in some applications, such as audio work, where a momentary open-circuit can introduce clicks and other undesirable responses. By using a narrow rotor-blade pickup, nonshorting designs break the previous circuit before making the next. Be careful to specify the version required for your circuit.

Another type of contact uses multiple-spring leaves. They are used in Edison Electronics Daven Series B selector switches and the Ross units. The spring leaves make contact with fixed studs. Such construction is usually used on large switches that can handle amperes of current. Or the leaves can provide low contact resistances in precision decade boxes or low-resistance bridges.

The third type of contact is used in printed-circuit configurations. It uses one or more "brushes" that ride on the surface of the contacts. This type, though, is used mainly in thumbwheel switches. The pressure between the contacts is necessarily light—60 to 70 grams for standard Chicago Dynamics' switches—to limit the wear on plated contacts and allow a light touch in thumbwheel switches. Too light a contact pressure will, of course, produce erratic results.

In addition to contact construction, conventional rotaries also offer a choice between open and enclosed construction. The lower-cost, open types are receiving rough competition from the



The color and PC compatibility of EECO's 800 series thumbwheel switches combine to make a compact traffic-control module.

enclosed. RCL says that it has reached the point where its commercial enclosed miniature switches are price-competitive with the older, open-frame wafer switches. RCL makes a complete line of enclosed rotaries—from 1/2-in. to 1-3/8-in. in diameter.

Some rotary switches have followed the trend to PC assembly techniques. Many traditional types have been redesigned to mount to PC boards, with terminations for automatic soldering. New switches have been specially configured for efficient PC mounting.

For example, Stackpole's Series 80 switch, announced for this fall, is rectangular. This shape allows all terminals to be in-line, with 0.1-in. centers. Special or BCD coding will be available with 30, 60 and 90-degree indexing, and up

to nine decks, with two modules per deck, will be offered.

Although we tend to think of rotary switches as wafers with shaft-driven wipers, there are other ways of constructing a rotary switch. In some the shaft rotates a permanent magnet, which actuates hermetically sealed reed switches—such as the Gordos GRS series, which is only 1 D × 1 H in. and weighs less than 1 oz.

Thumbwheels are more than a sidewise rotary

When switch designers first thought of turning a conventional rotary switch sidewise and then added a thumbwheel actuator, they got more than they expected. A thumbwheel switch offers minimum panel space, highly readable and repeatable display; rapid low-fatigue switching; the ability to control complex circuits, and coding and decoding with a minimum of wiring.

The most important asset of a single thumbwheel is its one-digit-at-a-time display. A bank of thumbwheel switches can provide almost any kind of data input required by a computer or machine control. A 10-digit number with the digits spaced only 3/8-in. apart—as produced by stacked-thumbwheel modules—is far easier to read than a display from rotaries that might need 1-in. spacing.

And, of course, color and lighting are used extensively to further enhance the readability and identification of the thumbwheel display. Blank spacers to fit the style of the switch and some spacers with markings—decimal points, symbols or abbreviations for minutes, seconds, etc.—easily fit into a switch assembly.

Sound and feel should be 'right'

A poorly designed switch scrunches into position with an annoying feel and sound. The feel and sound should be positive, distinct and pleasant. Human-factors engineers know that a reassuring audible click is needed to tell the user that the switch has snapped into place. Even when the construction of the switch does not require it, an audible snap is often built in by manufacturers to fill this psychological need.

In rotary switches, however, a positive detent is necessary for the switch's reliability. A good detent prevents the operator from teasing the switch and possibly setting it between valid positions.

Most thumbwheels require only 7 to 10 oz-in. of torque, in contrast with the 20 oz-in. and up for conventional rotaries. The touch of the thumb is lighter than the relatively powerful force of the wrist on a large knob.

But force is not the only measure of a good indexing mechanism. The shape of the thumbwheel is also important. Because the thumbwheel is used for frequent and rapid programming, the operating space on the wheel must fit the average operator's finger, provide a good, slip-free grip and avoid sharp tabs that can cause pain and fatigue. Thin, serrated edges wear on the operator's thumb and can be painful with constant use. Thumb-actuating wheels should be wide enough to fit most of an average thumb pad.

Operating with a gloved hand

Many applications—in military aircraft, industrial environments and elsewhere—require the operator to wear gloves. For these applications, switch designers have come up with so-called thumbwheel versions that use levers, rockers and pushbuttons to move the dial back and forth.

The pushbutton is easy to locate and operate with gloves, but the push-release action required is a bit slower than that for a straight thumbwheel. Chicago Dynamic's Series MPB/AS 27000 switches, Digitran's Series 12000 and Alco's MICO/MHE series are examples of pushbutton switches.

To improve the speed of dial setting, you can use lever actuators. In most lever-actuated rotaries, a 90-degree lever motion translates into a 360-degree motion with a rack-and-pinion assembly. This gear-up action allows rapid setting and resetting of the switch. With a light touch by a gloved hand, an entire bank of switches can be reset to zero with one motion. Digitran's Series 28000 Minilever and Cherry's Series L Leverwheel types are examples of lever-actuated "thumbwheel" rotaries.

Oak makes a rocker switch with an actuator that is a cross between a lever and a pushbutton. However, the indexing action is like that of a pushbutton—step by step. Two rockers provide bidirectional motion. Digitran's Series 24000 is, however, actuated by a single rocker lever that has a mechanism for one-digit-at-a-time action. Here a single lever provides bidirectional operation, and it is mounted below the display, so the operator's finger does not obscure the number display.

But the lever presents its own set of human-engineering problems. If it is too long, it can interfere with the display's visibility. If too short, it's hard to grip, especially with gloves. In either case, the safety advantage of rotary switches is negated. The switch position is vulnerable to accidental change, like a toggle switch. ■■

(continued on page 66)

Need more information?

We wish to thank the companies that provided information for this report. The products cited in the report have been selected for their illustrative, or in some cases, unique qualities. However, manufacturers not mentioned in the report may offer similar products. Readers may wish to consult manufacturers listed here for further details. Code letters listed after each company refer to switch types manufactured: rotary (R), thumbwheel (T).

AMP Inc., 449 Eisenhower Blvd., Harrisburg, Pa. 17105 (717) 564-0101. (R) (T) **Circle 400**

ASM Inc., 525 Truck Lane, Smithfield, N.C. 27577. (919) 934-3104. (R) **Circle 401**

Adgraphix Corp., 5940 Kester Ave., Van Nuys, Calif. 91401. (213) 781-1100. (T) **Circle 402**

Airflyte Electronics Co., New Hook Rd., Bayonne, N.J. 07002. (201) 436-2230. (R) **Circle 403**

Alco Electronic Products Inc., 1551 Osgood St., North Andover, Mass. 01845. (617) 685-4371. (R) (T) **Circle 404**

Allen-Bradley Co., 1201 South 2 St., Milwaukee, Wis. 53204. (414) 671-2000. (R) **Circle 405**

American Solenoid Co. Inc., 245 E. Inman Ave., Rahway, N.J. 07065. (201) 381-5100. (R) **Circle 406**

Amphenol RF Div. Bunker-Ramo Corp., 33 E. Franklin St., Danbury, Conn. 06810. (203) 743-9277. (R) **Circle 407**

Arrow-Hart, Inc., 103 Hawthorn St., Hartford, Conn. 06106. (203) 249-8471. (R) **Circle 408**

Astrosystems Inc., Nevada Dr., Lake Success, N.Y. 11040. (516) 328-1600. (R) (T) **Circle 409**

Beckman Instruments Inc., 2500 Harbor Blvd., Fullerton, Calif. 92634. (714) 871-4848. (R) (T) **Circle 410**

Bendix, Electrical Components Div., Sidney, N.Y. 13838. (607) 563-9511. (R) **Circle 411**

Bowmar/TIC Inc., Bowmar Inst. Corp., 850 Lawrence Dr., Newbury Park, Calif. 91320. (805) 498-2161. (R) **Circle 412**

C&K Components Inc., 103 Morse St., Watertown, Mass. 02172. (617) 926-0800. (R) **Circle 413**

CTS Keene Inc., 3230 Riverside Ave., Paso Robles, Calif. 93446. (805) 238-0350. (R) **Circle 414**

Carlingswitch Inc., 1428 Brickell Ave., Suite 800, Miami, Fla. 33131. (305) 358-5400. (R) **Circle 415**

Carson Mfg. Co. Inc., 5453 N. Rural, Indianapolis, Ind. 46220. (317) 257-3191. (R) **Circle 416**

Carter Mfg. Corp., 23 Washington St., Hudson, Mass. 01749. (617) 562-6987. (R) **Circle 417**

Centralab Div., Globe-Union Inc., 5757 N. Greenbay Ave., Milwaukee, Wis. 53201. (414) 228-1200. (R) **Circle 418**

Cetec Inc., 13035 Saticoy St., North Hollywood, Calif. 91605. (213) 875-1900. (R) **Circle 419**

Cherry Electrical Products Corp., Waukegan-Gurnee Ind. Park, Waukegan, Ill. 60085. (312) 689-7600. (T) **Circle 420**

Chicago Dynamic Industries Inc., Precision Products Div., 1725 W. Diversey Blvd., Chicago, Ill. 60614. (312) 935-4600. (R) (T) **Circle 421**

Chicago Lock Co., 4275 W. Belmont Ave., Chicago, Ill. 60641. (312) 282-7177. (R) **Circle 422**

Chicago Switch Inc., 2035 W. Wabansia, Chicago, Ill. 60647. (312) 489-5500. (R) (T) **Circle 423**

Cole-Hersee Co., 20 Old Colony Ave., Boston, Mass. 02127. (617) 268-2190. (R) **Circle 424**

Cole Instrument Corp., 2034 Placentia Ave., Costa Mesa, Calif. 92627. (714) 642-8080. (R) **Circle 425**

Connector Corp., 6025 N. Keystone, Chicago, Ill. 60646. (312) 539-3108. (R) **Circle 426**

Couch S. H. Div. 36 River St., Boston, Mass. 02126. (617) 298-4147. (R) **Circle 427**

Cutler-Hammer Inc., 4201 North 27th St., Milwaukee, Wis. 53216. (414) 442-7800. (R) **Circle 428**

Dension, R. B. Inc., 103 Broadway, Bedford, Ohio 44146. (216) 232-8200. (R) **Circle 429**

Dialight Corp., 60 Stewart Ave., Brooklyn, N.Y. 11237. (212) 497-7600. (T) **Circle 430**

Digitran Co., 855 S. Arroyo Parkway, Pasadena, Calif. 91105. (213) 449-3110. (T) **Circle 431**

Disc Instruments, Inc., 102 E. Baker St., Costa Mesa, Calif. 92626. (714) 979-5300. (R) **Circle 432**

Duncan Electronics, 2865 Fairview Rd., Costa Mesa, Calif. 92626. (714) 545-8261. (R) **Circle 433**

Durant Digital Instruments Div., Cutler-Hammer, 901 South 12th St., Watertown, Wis. 53094. (414) 261-4070 (T) **Circle 434**

(EECO) Electronic Engrg. Co. of California, Electronic Products Div., 1441 E. Chestnut Ave., Santa Ana, Calif. 92701. (714) 835-6000. (R) (T) **Circle 435**

Eagle Electric Mfg. Co., 45-31 Court Square, Long Island City, N.Y. 11101. (212) 937-8000. (R) (T) **Circle 436**

Edison Electronics Div., McGraw-Edison Co., Grenier Field Municipal Airport, Manchester, N.H. 03103. (603) 669-0940. (R) **Circle 437**

Electro-Mec Instrument Corp., 382 Main St., Yalesville, Conn. 06492. (203) 269-7711. (R) **Circle 438**

Electro Switch Corp., King Ave., Weymouth, Mass. 02188. (617) 335-5200. (R) **Circle 439**

Essex, 1601 Wall St., Fort Wayne, Ind. 46804. (219) 743-0311. (R) **Circle 440**

Fisher Controls Co., 205 S. Center St., Marshalltown, La. 50158. (515) 754-3011. (R) **Circle 441**

Full Linear Control Co., 417 Liberty St., Jackson, Mich. 49201. (517) 783-5343. (R) **Circle 442**

Furnas Elec. Co., 1197 McKee St., Batavia, Ill. 60610. (312) 879-3410. (R) **Circle 443**

GC Electronics, Div., Hydrometals Inc., 400 S. Wyman St., Rockford, Ill. 61101. (815) 968-9661. (R) **Circle 444**

General Control Co., 1200 Soldiers Field Rd., Boston, Mass. 02134. (617) 782-7440. (R) **Circle 445**

General Devices Inc., Instrumentation Div., P.O. Box 253, Princeton, N.J. 08540. (609) 924-2500. (R) **Circle 446**

General Elec. Co., 1 River Rd., Schenectady, N.Y. 12345. (518) 374-2211. (R) **Circle 447**

General Microwave Corp., 155 Marine St., Farmingdale, N.Y. 11735. (516) 694-3600. (R) **Circle 448**

Gordos Corp., 250 Glenwood Ave., Bloomfield, N.J. 07003. (201) 743-6800. (R) **Circle 449**

Grayhill Inc., 545 Hillgrove Ave., La Grange, Ill. 60525. (312) 354-1040. (R) **Circle 450**

Gunther America Inc., 9 Manee Rd., Holmdel, N.J. 07733. (201) 264-2488. (R) **Circle 451**

Hathaway Industries Inc., P.O. Box 45381, Tulsa, Okla. 74145. (918) 663-0110. (R) **Circle 452**

Haydon Switch & Instrument Inc., 1500 Meriden Rd., Waterbury, Conn. 06720. (203) 756-7441. (R) **Circle 453**

ITT Components Group, 1551 Osgood St. N. Andover, Mass. 01845. (617) 688-1881. (R) (T) **Circle 454**

Imlac Corp., 150 A St., Needham, Mass. 02194. (617) 449-4000. (T) **Circle 455**

Impact Electrical Products Inc., 333 Adams St., Bedford Hills, N.Y. 10507. (914) 666-8939. (R) **Circle 456**

Imtronics Ind. Ltd., 48 Commerce Dr., Farmingdale, N.Y. 11735. (516) 293-5282. (R) **Circle 457**

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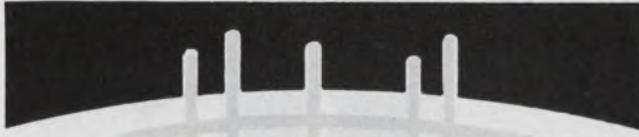
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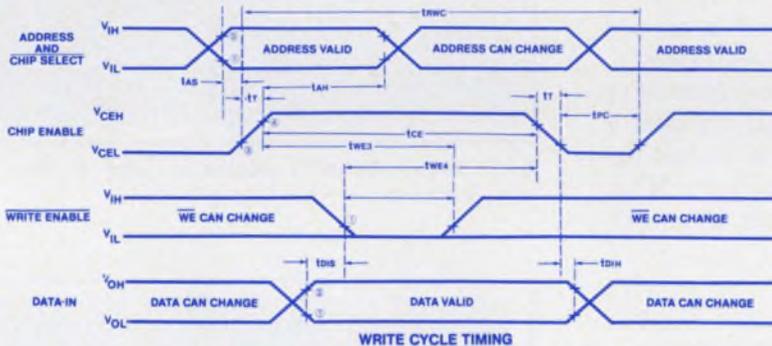
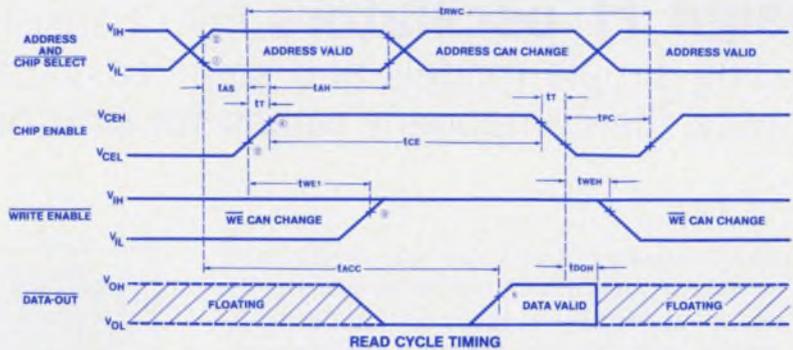


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INFORMATION RETRIEVAL NUMBER 47

Design rf oscillators from transistor data sheets.

By using simple models and some easy math, you can compute the component values for sinusoidal oscillators.

Analyze the data on transistor spec sheets, and you'll be able to simplify rf oscillator design. By using the manufacturer's data and simple transistor models, you can design stable oscillators in the 1-to-100-MHz frequency range.

Many different circuit configurations are possible, but let's consider the Colpitt's oscillator with a sinusoidal output. This type of oscillator uses a tuned circuit but doesn't need a tapped inductor, as required in the Hartley oscillator.

If all designs are limited to the low-frequency region—say, below 1 MHz—internal transistor capacitances have very little effect on circuit operation. But as frequencies increase above 1 MHz, the internal effects play havoc with the basic equation for oscillator frequency:

$$f = 1 / 2\pi\sqrt{LC}.$$

In this equation L and C represent the discrete values of tank inductance and capacitance.

A look at the Colpitt's oscillator

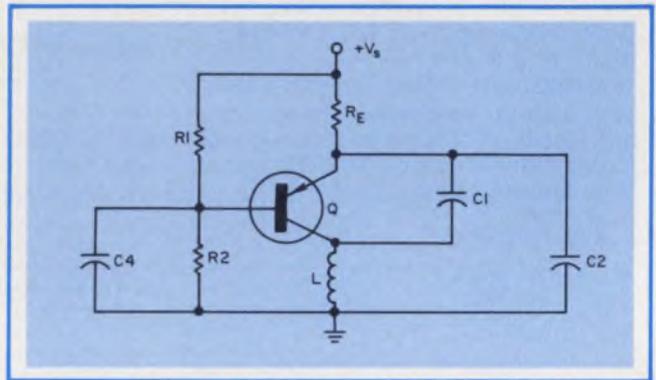
The basic Colpitt's oscillator (Fig. 1) provides sinusoidal outputs over a wide range of frequencies. As a typical example, let's design a circuit that provides sinusoidal oscillations at 30 MHz.

To start the design, the π model for the pnp transistor can be very useful when you analyze the circuit. Since the oscillator circuit uses a transistor with a grounded base (C_1 is an ac short at 30 MHz), a simplified model can be used. Also, the output and feedback internal resistances are usually much greater than any load resistance that would be used for an oscillator and can thus be neglected. The simplified π model,¹ with appropriate conversion factors considered, minimizes calculations (Fig. 2) to determine component values.

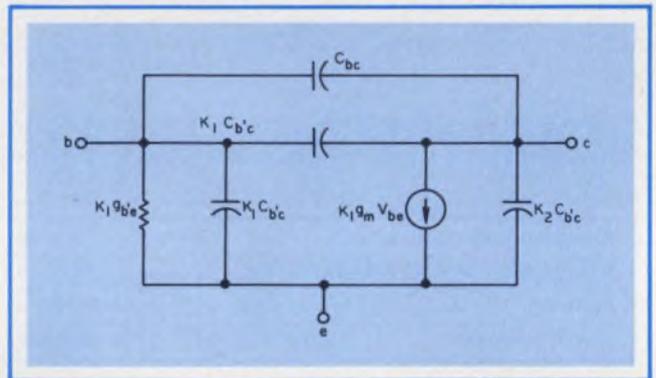
The model is valid over a frequency range that has an upper limit² determined by

$$(f_c/3)(r_e/r_{bb'}).$$

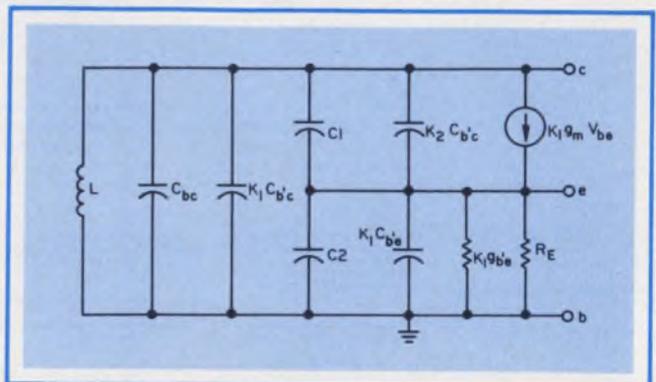
If the external components shown in the schematic of Fig. 1 are added to the equivalent circuit (also allowing for C_1 as an ac short at 30



1. A simple Colpitt's oscillator uses a pnp transistor and produces sine-wave oscillations.

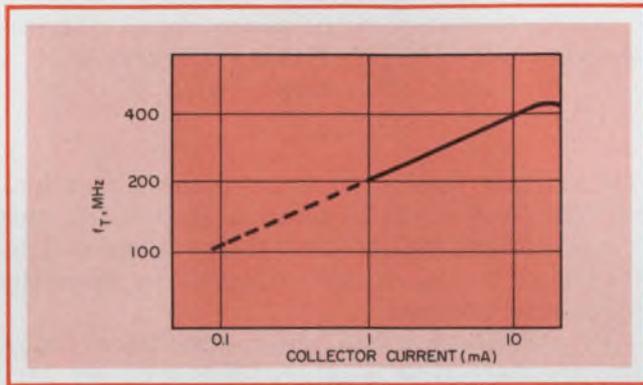


2. The simple pnp transistor π model can help reduce the number of calculations when you compute the values of the internal capacitances.

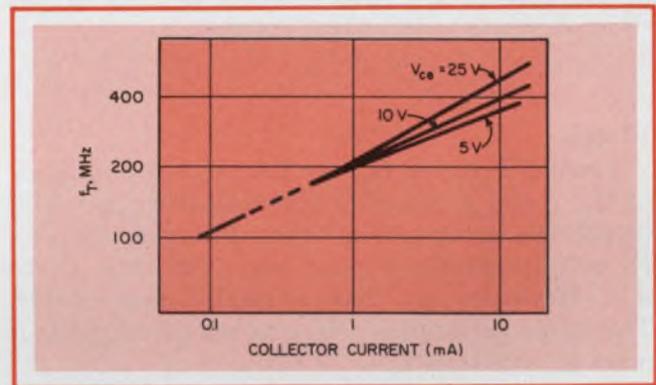


3. If the external components are added to the π model, the oscillator circuit becomes ready for analysis.

Fred Palenschat, Chief Electrical Engineer, Gamma Scientific, 3777 Ruffin Rd., San Diego, Calif. 92123



4. The transistor f_t decreases almost linearly with the decreasing collector current.



5. As the collector-emitter voltage varies over a 5-to-25-V range, f_t holds almost constant.

MHz), the schematic in Fig. 3 results. Resistors R_1 and R_2 are also neglected.

To ease circuit analysis, some of the parameters in the diagram of Fig. 3 can be defined (Table 1). All have some control over the capacitance that affects the oscillator performance.

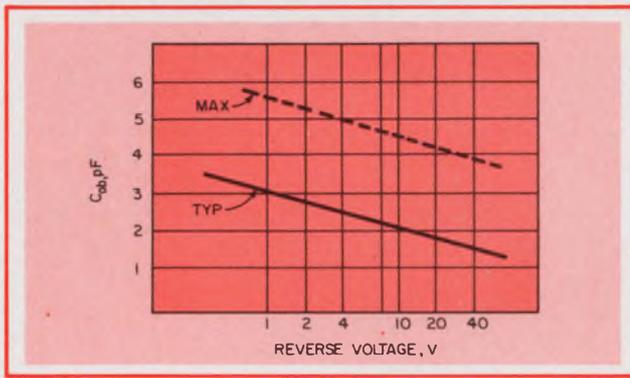
From Table 1, it is obvious that g_m is directly proportional to the emitter current ($r_e \approx 26 \text{ mV}/I_E$). Since $C_{b'e}$ is itself directly proportional to g_m and inversely proportional to f_t , either I_E must

be kept fairly low or the transistor's f_t must be high. Usually a compromise between f_t and I_E is made (Fig. 4), but this causes complications, since f_t decreases as the emitter current decreases. Over the same current range as in Fig. 4, f_t will remain nearly constant with variations in collector-emitter voltage (Fig. 5).

As a rule of thumb (if no curves are supplied on the data sheet), f_t is approximately halved for every decade of decrease in emitter current below

Table 1. Parameter definitions for transistor model

| | |
|--|--|
| $g_m = \alpha/r_e$ | where α (or $\beta/(1 + \beta)$) is the low frequency common-base current gain, and r_e is the dynamic resistance looking in at the emitter. |
| $g_{b'e} = g_m/\beta$ | where β is the dc common emitter current gain. |
| $g_{bb'} = 1/r_{bb'}$ | where $r_{bb'}$ is the base spreading resistance. |
| $C_{b'c} = C_{ob} - C_{bc}$ | where C_{bc} is the header capacitance and overlap diode capacitance ³ , and C_{ob} is the collector-base output capacitance with the emitter open (usually specified). |
| $C_{b'e} = (g_m/2\pi f_t) - C_{b'c}$ | where f_t is the frequency at which $\beta = 1$ (the gain-bandwidth-product). |
| $K_1 = g_{bb'}/(g_{bb'} + g_{b'e})$ $K_2 = (g_m + g_{b'e})/(g_{bb'} + g_{b'e})$ | |



6. The value of C_{ob} varies inversely with the applied value of reverse collector-base voltage.

10 mA.

From Fig. 3, you can see that K_2 is a multiplier of $C_{b'e}$ —and a potential problem. Since $g_{b'e}$ will usually be much smaller than either g_m or $g_{bb'}$, K_2 will essentially depend upon the ratio g_m to $g_{bb'}$. However, $g_{bb'}$ has a fairly stable value. Therefore variations in K_2 are heavily dependent upon g_m and thus I_E .

Select the right transistor

From the analysis so far the most important factors to consider in transistor selection are the f_t and the emitter current. Note, though, in Fig. 6 that $C_{b'e}$, and thus C_{ob} , is fairly independent of current but varies inversely with reverse voltage.⁵

For the design example, a 2N4916 transistor was chosen because of its high f_t . However, the manufacturer's data sheet notes that the gain-bandwidth curves are typical composites. Thus to design for worst-case conditions, you can find the minimum f_t by first checking the minimum high-frequency value of h_{fe} on the data sheet. For the specific transistor selected, the h_{fe} minimum at 100 MHz is 4. Thus the gain-bandwidth product is 400 MHz.

Some of the other parameter values that will be

Table 2. Selected transistor parameters

| |
|--|
| $h_{FE}(\text{min}) = 50$ |
| $h_{fe}(\text{min}) = 4$ at $I_E = 10$ mA, $V_{CE} = -20$ V, and $f = 100$ MHz |
| $\therefore f_t(\text{min}) = 400$ MHz at $I_E = 10$ mA and $V_{CE} = -20$ V |
| $C_{ob}(\text{max}) = 4.5$ pF at $I_E = 0$, and $V_{cb} = -10$ V |
| $r_{b'e} = 50$ ps at $I_c = 10$ mA, and $V_{CE} = -20$ V |

used in the design example are shown in Table 2. The calculations can proceed if a supply voltage of 9 V and a V_{CE} drop of 5 V can be assumed. The chosen V_{CE} is quite arbitrary, but it was chosen to be high enough to prevent gross degradation of f_t .

From Table 2, you are given C_{ob} as equal to 4.5 pF at a -10 -V bias. And from the data sheet, you can see that C_{ob} increases from 4.5 to 4.8 pF as the collector voltage is reduced to 5 V (Fig. 6). If a header capacitance and overlap diode capacitance of 2 pF is assumed, you can find that

$$C_{b'e} = 4.8 - 2 \text{ pF} = 2.8 \text{ pF.}$$

As the first step in the design, let $C_{o'e}$ be less than or equal to 45 pF (this is arbitrary). But

$$C_{b'e} = (g_m/2\pi f_t) - C_{b'e}$$

$$\therefore C_{b'e} = (\alpha I_E/52\pi f_t) - C_{b'e}$$

$$\text{and } I_E = (C_{b'e} + C_{b'e}) (52\pi f_t/\alpha).$$

The original value of f_t was measured at a bias of -20 V. Very little error is incurred if this value is used; typical data sheet values will be used later in the design procedure to show the differences obtained.

By substituting data sheet parameter values into the theoretical equation for I_E , you get

$$I_E = 3.174 \text{ mA.}$$

Since the calculated emitter current is closer to 1 mA, a lower f_t should be used—say, 200 MHz. Now if you recalculate the emitter current, you get $I_E = 1.587$ mA.

Determine the bias resistors

Once you know the emitter current, the base bias network can be designed. Start by redrawing the base circuit in Fig. 1 to that of Fig. 7. Let S , the stability factor, be equal to 2.2 (again quite arbitrary; good bias stability dictates that S should be between 2 and 10).

Since $S = (R_B/R_E) + 1$, the ratio of R_B to R_E can be computed to be 1.2. The V_{CE} was already stipulated at 5 V. Thus the value of R_E becomes $(V_{\text{supply}} - V_{CE})/I_E$ or 4 V/1.587 mA, which equals 2.5 k Ω . To simplify the design, let $R_E = 2.2$ k Ω , a standard component value. Since R_E is now known, R_B can be found:

$$R_B = 1.2 (2.2 \text{ k}\Omega) = 2.64 \text{ k}\Omega$$

From the circuit of Fig. 7, V_{th} can be calculated:

$$V_{th} = [I_E \cdot (1 + \beta)] [R_B + (1 + \beta) R_E] + 0.6 \text{ V} \\ = 4.16 \text{ V.}$$

Now the values of R_1 and R_2 can be calculated:

$$R_B = R_1 // R_2 \text{ and } (V_{\text{supply}}) \left(\frac{R_2}{R_1 + R_2} \right) = V_{th}$$

thus $R_1 = 5.71$ k Ω and $R_2 = 4.91$ k Ω .

If you let these resistors take on standard values, you get $R_1 = 5.6$ k Ω and $R_2 = 4.7$ k Ω . Using these new values, you can recalculate the value of I_E to find $I_E = 1.564$ mA.

Once all these component values have been

found, the transistor model parameters can be determined:

$$g_m = (1.564) (60) / (260) (61) = 59.17 \text{ mmhos},$$

$$g_{b'e} = g_m / \beta = 0.986 \text{ mmhos},$$

$$r_{bb'} = r_b C_c (\text{max}) / C_{b'e} (\text{max}),$$

with $r_b C_c$ given at 10 mA and -20 V. Capacitance C_{ob} is relatively constant with current but is reduced at 20 V to 4.2 pF (from manufacturer's data sheet).

Thus $C_{b'e} = 4.2 \text{ pF} - 2 \text{ pF} = 2.2 \text{ pF}$ and, in turn, $g_{bb'} = 2.2 \text{ pF} / 50 \text{ ps} = 44 \text{ mmhos}$. From the formulas in Table 1, you can now find the values of K_1 , K_2 , $C_{b'e}$, C_{bc} and $C_{b'c}$. Their values are 0.978, 1.337, 2.8 pF, 2 pF and 44.3 pF, respectively.

All these calculated values can now be substituted into the modeled transistor circuit of Fig. 3 (redrawn in Fig. 8). You can also select arbitrary values of C_1 and C_2 , although C_2 is normally kept between the limits of $C_1 \leq C_2 \leq 3C_1$. Also, the values of C_1 and C_2 are chosen so as not to mask or swamp out the internal stray capacitance (to help illustrate this design procedure).

You can now compute the total capacitance by adding up the values of C_{bc} , $K_1 C_{b'e}$ and allowing for the series-parallel arrangement of C_1 , C_2 , $K_2 C_{b'e}$ and $K_1 C_{b'e}$. Thus you get

$$C_t = C_{bc} + K_1 C_{b'e} + \frac{(C_1 + K_2 C_{b'e})(C_2 + K_1 C_{b'e})}{C_1 + K_2 C_{b'e} + C_2 + K_1 C_{b'e}}$$

$$C_t = 15 \text{ pF}.$$

From the basic formula for oscillation, you can substitute in the value of C and find the value of inductance needed for oscillation at a frequency of 30 MHz:

$$L = 1 / (2\pi \times 30 \times 10^6)^2 C_t = 1.88 \mu\text{H}.$$

As a check on the accuracy of the method, assume that a 2.5- μH choke is used and the frequency recalculated and measured. The calculated worst-case frequency becomes 26 MHz and the measured frequency is 34.5 MHz.

Remember, the external capacitors should be selected so that internal capacitances in the final computations will have little or no effect on the oscillator frequency. This is not the case in the example so far—there is a wide disagreement between the calculated and measured frequencies.

The procedure can be further improved

If the emitter current in the transistor is reduced by an order of magnitude, you can get closer agreement in the frequency. For example, when R_E increases to 33 k Ω , I_E drops to 0.106 mA and β decreases by 30%. The other parameter values then change to the following:

$$g_m = 3.977 \text{ mmhos} \quad K_2 = 0.09$$

$$g_{b'e} = 0.099 \text{ mmhos} \quad C_{b'e} = 2.8 \text{ pF}$$

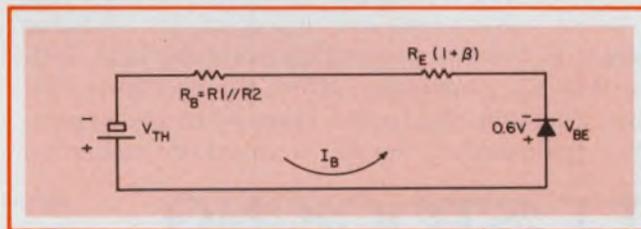
$$g_{bb'} = 44 \text{ mmhos} \quad C_{bc} = 2 \text{ pF}$$

$$K_1 = 1$$

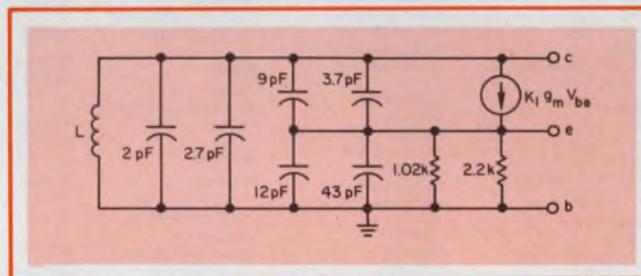
Also, for the decrease in current, f_i drops by

Table 3. Comparison of computed oscillator frequencies

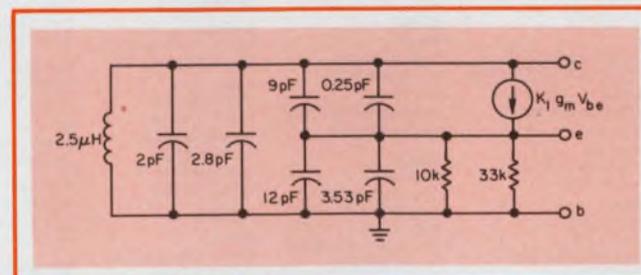
| $I_E = 1.564 \text{ mA}$ | $I_E = 0.106 \text{ mA}$ |
|-------------------------------------|-------------------------------------|
| $h_{FE} = 100$ | $h_{FE} = 70$ |
| $f_T = 300 \text{ MHz}$ | $f_T = 150 \text{ MHz}$ |
| $C_{ob} = 2.5 \text{ pF}$ | $C_{ob} = 2.5 \text{ pF}$ |
| $r_{bb'} = 23 \Omega$ | $r_{bb'} = 23 \Omega$ |
| $g_{bb'} = 44 \text{ mmhos}$ | $g_{bb'} = 44 \text{ mmhos}$ |
| $g_m = 51.6 \text{ mmhos}$ | $g_m = 3.48 \text{ mmhos}$ |
| $g_{b'e} = 0.516 \text{ mmhos}$ | $g_{b'e} = 0.0498 \text{ mmhos}$ |
| $C_{b'e} = 0.5 \text{ pF}$ | $C_{b'e} = 0.5 \text{ pF}$ |
| $C_{bc} = 2 \text{ pF}$ | $C_{bc} = 2 \text{ pF}$ |
| $C_{b'c} = 26.9 \text{ pF}$ | $C_{b'c} = 3.2 \text{ pF}$ |
| $K_1 = 1$ | $K_1 = 1$ |
| $K_2 = 1.17$ | $K_2 = 0.081$ |
| $f(\text{calc}) = 31.5 \text{ MHz}$ | $f(\text{calc}) = 35.2 \text{ MHz}$ |
| $f(\text{meas}) = 34.5 \text{ MHz}$ | $f(\text{meas}) = 35.2 \text{ MHz}$ |



7. The simplified base circuit of the transistor can be used to design the bias network.



8. The completely calculated transistor model can now be simplified so that the inductor value can be found.



9. Recalculating the transistor model components with typical values from the data sheet, you can find the value of the tank inductor.

Table 4. Worst-case & typical vs measured frequencies

| I_E (mA) | C_1 (pF) | C_2 (pF) | L (μ H) | f(calc) worst case (MHz) | f(calc) (MHz) | f(meas) (MHz) |
|------------|------------|------------|----------------|--------------------------|---------------|---------------|
| 1.564 | 29 | 29 | 0.53 | 41.9 | 46.7 | 50.8 |
| 0.106 | 29 | 29 | 0.53 | 48.6 | 51.9 | 51.3 |
| 1.564 | 29 | 100 | 0.53 | 39 | 42.6 | 40.8 |
| 0.106 | 29 | 100 | 0.53 | 41.5 | 43.5 | 43.5 |
| 1.564 | 9 | 12 | 2.5 | 26 | 31.5 | 34.5 |
| 0.106 | 9 | 12 | 2.5 | 30.9 | 35.2 | 35.2 |

a factor of 2; thus $f_t = 100$ MHz. And $C_{b'e}$ then becomes 3.53 pF. All these values can now be substituted into the same circuit as before, and you can recompute and measure the frequency of oscillation. The values become 30.9 MHz (calculated) and 35.2 MHz (measured). The measured and calculated values are much closer than before, since C_1 and C_2 are more dominant.

This mathematical approach not only is in close agreement but also gives a strong indication as to which parameters are important in oscillator design. In the two preceding examples it is obvious that $C_{b'e}$ variations affect the oscillator frequency less in the second case, since the capacitance calculation shows a marked reduction

(more than an order of magnitude). Therefore good design requires that C_1 and C_2 be selected to "swamp out," or at least be a large part of their parallel stray counterparts.

Let's do the same design example once more, using the typical parameters from the transistor data sheet. Table 3 has a comparison of all the calculated and given parameters for the two previous emitter-current levels.

To show the over-all effect of tank-capacitance selection, only the tank components have been changed, and the frequency again calculated and measured. Table 4 lists the previous examples as well as some other arbitrary values for comparison.

Thus if C_1 and C_2 are indeed selected to mask the inherent stray capacitances—say, $C_2 = 2C_1$, $C_2 > 2K_1C_{b'e}$, $C_1 \geq 3C_{ob}$ and $C_1 > 5C_{b'c}$ —the calculated and the measured values will agree. Even if the worst-case parameters at a given temperature are used, the results are still usable. ■■

References

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2. Ibid p. 57.
3. Searle, Boothroyd, Angelo, Grey, and Pederson, "Elementary Circuit Properties of Transistors," S.E.E.C., Wiley, 1964, Vol. 3; p. 102.
4. Millman and Halkias, "Integrated Electronics," McGraw-Hill, 1972, p. 358.
5. Data Sheet 2N4916; Fairchild Semiconductor.

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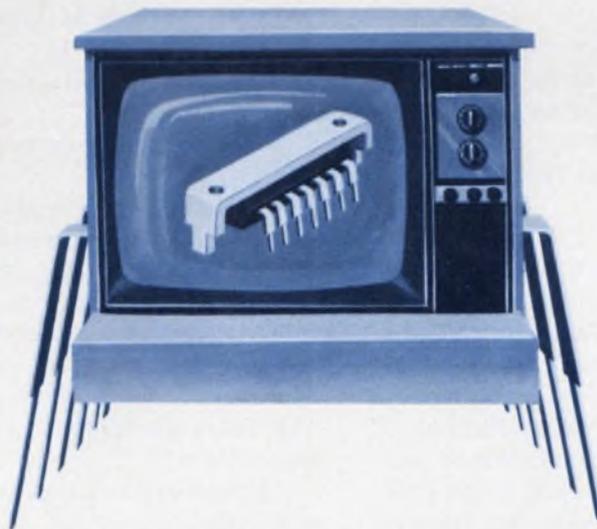
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Convert two-port circuit parameters

from any form to any other. A short Fortran program handles all popular parameters in polar or rectangular form.

Two-port parameter conversions are a time-consuming task to program or calculate. But the 140 equations required for these conversions can be reduced to less than 100 Fortran statements, if you take advantage of the redundancies inherent in the equations.

A compact Fortran program results, which performs all possible conversions between any pair of six two-port parameter systems. The user can input the four matrix elements for any of the six— z , y , h , g , ABCD and S —and select conversion to any other form. The subroutine also carries out impedance normalization for S -parameters, provides output in rectangular as well as polar form, and accepts inputs in either rectangular or polar form.

Coding reduces program size

A coding scheme reduces the number of expressions programmed from 144 to 27 in the case of conversion, and from 24 to 7 for S -parameter normalization.

Conversion from one parameter matrix to another requires four equations—aside from possible normalizations. If we describe the input matrix in terms of four elements x_{11} , x_{12} , x_{21} and x_{22} , then the output matrix, R , has four elements. Each is given by an equation

$$r_{ij} = f_{ij}(x_{11}, x_{12}, x_{21}, x_{22}) \quad (1)$$

in which indices i and j range from one to two.

These functions f_{ij} contain considerable redundancy. Many simply require division by the determinant of the input matrix—regardless of which r_{ij} is to be calculated. And the number of possible numerator expressions is also limited.

All possible numerator and denominator expressions are calculated once and assigned a code number (Table 1). Then in each case, one item is selected for the numerator, another for the denominator, and the ratio, r_{ij} , is computed.

For simplicity, the actual program performs all 27 calculations on the input variables, then

stores them in array X with the subscript equal to the code number.

Five matrices of code numbers are formed to reference the pre-stored conversion computations (Table 2). The matrices N_{11} , N_{12} , N_{21} and N_{22} contain the locations of numerator terms for elements r_{11} , r_{12} , r_{21} and r_{22} . The common-denominator term location is stored in array D .

The use of the arrays is straightforward. For instance, to compute the element y_{12} in a S -to- y conversion, matrix N_{12} indicates the numerator code 30 and matrix D indicates the denominator code 13. And from Table 1, these codes show that

$$y_{12} = \frac{-S_{12}}{S_{11}S_{12} - S_{12}S_{21} + S_{11} + S_{22} + 1} = \frac{X(30)}{D(13)} \quad (2)$$

The computations for S -parameter normalization formulae are stored in array N , and the multiplicand or divisor is chosen for each of the four elements from array $NORM$ (Table 3).

How to use the program

Each problem is entered with the following information (Fig. 1):

- Numerical designation of input-parameter type.
- Numerical designation of output-parameter type.
- Polar or rectangular designation on input.
- Values for impedance-level normalizations on input and output—if S -parameters are specified for input or output.
- Appropriate numerical values for the four input elements (R , θ) or (x , y).

Data calculated and outputted are the values of the four parameters in both rectangular and polar form.

The program package consists of a mainline and subroutines AZCON, RECT and POLAR. AZCON performs all conversions, including the normalization/denormalization. RECT and POLAR provide polar-to-rectangular and rectangular-to-polar conversions. These function routines are used only by AZCON. The mainline program handles input and output to the user and passes parameters to AZCON.

Once AZCON, RECT and POLAR have been

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```

INPUT,OUTPUT (1=Z,2=Y,3=H,4=G,5=ABCD,6=S) ? 6 2
INPUT FORM(1=RECT,2=POLAR) ? 2
ROI,RO2=? (DEFAULT=50) ?
S11 S21 S12 S22
? .35 259,2.5 78,.092 56,.68 337
IF NORMALIZATION REQUIREMENTS
ARE OTHER THAN 50Ω TYPE IN
DESIRED INPUT AND OUTPUT
IMPEDANCES.
NUMBERS ARE IN POLAR
NOTATION
REAL IMAGINARY MAGNITUDE ANGLE
Y11= 0.1280E-01 0.1340E-01 Y11= 0.1853E-01 46.32
Y21= 0.1795E-01 -.5068E-01 Y21= 0.5376E-01 -70.49
Y12= -.8610E-04 -.1977E-02 Y12= 0.1978E-02 -92.49
Y22= 0.9715E-03 0.4188E-02 Y22= 0.4300E-02 76.94

```

(Y)

Table 1. Code numbers assigned to conversion expressions

| Code No. | Computation |
|----------|--|
| 0 | None |
| 1 | X_{11} (A) |
| 2 | X_{21} (C) |
| 3 | X_{12} (B) |
| 4 | X_{22} (D) |
| 5 | $\Delta_x = X_{11} X_{22} - X_{12} X_{21}$ |
| 6 | 1 |
| 7 | $\Delta_x + X_{11} - X_{22} - 1$ |
| 8 | $\Delta_x - X_{11} + X_{22} - 1$ |
| 9 | $2 X_{21}$ |
| 10 | $2 X_{12}$ |
| 11 | $2 \Delta_x$ |
| 12 | 2 |
| 13 | $\Delta_x + X_{11} + X_{22} + 1$ |
| 14 | $\Delta_x - X_{11} - X_{22} + 1$ |
| 15 | $X_{11} + X_{12} + X_{21} + X_{22}$ |
| 16 | $X_{11} + X_{12} - X_{21} - X_{22}$ |
| 17 | $-X_{11} + X_{12} - X_{21} + X_{22}$ |
| 18 | Not used |
| 19 | Not used |
| 20 | Not used |
| 21 | $-X_{11}$ |
| 22 | $-X_{21}$ |
| 23 | $-X_{12}$ |
| 24 | $-X_{22}$ |
| 25 | $-\Delta_x$ |
| 26 | -1 |
| 27 | $-\Delta_x - X_{11} + X_{22} + 1$ |
| 28 | $-\Delta_x + X_{11} - X_{22} + 1$ |
| 29 | $-2 X_{21}$ |
| 30 | $-X_{12}$ |

Note: These computations on the 4-input elements x_{ij} are stored by code number in the 30-element array X.

```

INPUT,OUTPUT (1=Z,2=Y,3=H,4=G,5=ABCD,6=S) ? 6 1
INPUT FORM(1=RECT,2=POLAR) ? 2
ROI,RO2=? (DEFAULT=50) ?
S11 S21 S12 S22
? .35 259,2.5 78,.092 56,.68 337
REAL IMAGINARY MAGNITUDE ANGLE
Z11= 36.05 11.46 Z11= 37.83 17.63
Z21= 302.8 33.4 Z21= 473.0 50.20
Z12= 15.34 8.225 Z12= 17.41 28.20
Z22= 158.9 -36.64 Z22= 163.0 -12.99

```

(Z)

```

INPUT,OUTPUT (1=Z,2=Y,3=H,4=G,5=ABCD,6=S) ? 6 3
INPUT FORM(1=RECT,2=POLAR) ? 2
ROI,RO2=? (DEFAULT=50) ?
S11 S21 S12 S22
? .35 259,2.5 78,.092 56,.68 337
REAL IMAGINARY MAGNITUDE ANGLE
H11= 37.27 -39.03 H11= 53.97 -46.32
H21= -1.309 -2.589 H21= 2.901 -116.8
H12= 0.8035E-01 0.7031E-01 H12= 0.1068 41.19
H22= 0.5977E-02 0.1378E-02 H22= 0.6134E-02 12.99

```

(H)

```

INPUT,OUTPUT (1=Z,2=Y,3=H,4=G,5=ABCD,6=S) ? 6 4
INPUT FORM(1=RECT,2=POLAR) ? 2
ROI,RO2=? (DEFAULT=50) ?
S11 S21 S12 S22
? .35 259,2.5 78,.092 56,.68 337
REAL IMAGINARY MAGNITUDE ANGLE
G11= 0.2519E-01 -.8008E-02 G11= 0.2644E-01 -17.63
G21= 10.54 6.730 G21= 12.50 32.56
G12= -.4524 -.8437E-01 G12= 0.4460 -169.4
G22= 52.55 -226.6 G22= 232.6 -76.94

```

(G)

```

INPUT,OUTPUT (1=Z,2=Y,3=H,4=G,5=ABCD,6=S) ? 6 5
INPUT FORM(1=RECT,2=POLAR) ? 2
ROI,RO2=? (DEFAULT=50) ?
S11 S21 S12 S22
? .35 259,2.5 78,.092 56,.68 337
REAL IMAGINARY MAGNITUDE ANGLE
A= 0.6740E-01 -.4305E-01 A= 0.7997E-01 -32.56
C= 0.1353E-02 -.1624E-02 C= 0.2114E-02 -50.20
B= -6.211 -17.53 B= 18.60 -109.5
D= 0.1555 -.3076 D= 0.3447 -63.19

```

(ABCD)

```

INPUT,OUTPUT (1=Z,2=Y,3=H,4=G,5=ABCD,6=S) ? 6 6
INPUT FORM(1=RECT,2=POLAR) ? 2
ROI,RO2=? (DEFAULT=50) ?
S11 S21 S12 S22
? .35 259,2.5 78,.092 56,.68 337
REAL IMAGINARY MAGNITUDE ANGLE
S11= -.6678E-01 -.3436 S11= 0.3500 -101.0
S21= 0.5198 2.445 S21= 2.500 78.00
S12= 0.5145E-01 0.7627E-01 S12= 0.9200E-01 56.00
S22= 0.6259 -.2657 S22= 0.6800 -23.00

```

(S)

1. You can perform any of six two-port conversions by supplying the underlined information to the program.

Table 2. Arrays used as indices to numerator and denominator terms

| Four Numerator Arrays | | | | | | |
|-----------------------|------|----|----|----|----|----|
| To | From | | | | | |
| | z | y | h | g | A | S |
| z | 0 | 4 | 5 | 6 | 1 | 28 |
| y | 4 | 0 | 6 | 5 | 4 | 27 |
| h | 5 | 6 | 0 | 4 | 3 | 13 |
| g | 6 | 5 | 4 | 0 | 2 | 14 |
| A | 1 | 24 | 25 | 6 | 0 | 28 |
| S | 7 | 27 | 7 | 27 | 16 | 0 |

N 11

| To | From | | | | | |
|----|------|----|----|----|----|----|
| | z | y | h | g | A | S |
| z | 0 | 23 | 3 | 23 | 5 | 10 |
| y | 23 | 0 | 23 | 3 | 25 | 30 |
| h | 3 | 23 | 0 | 23 | 5 | 10 |
| g | 23 | 3 | 23 | 0 | 25 | 30 |
| A | 5 | 26 | 21 | 4 | 0 | 13 |
| S | 10 | 30 | 10 | 30 | 11 | 0 |

N 12

| To | From | | | | | |
|----|------|----|----|----|----|----|
| | z | y | h | g | A | S |
| z | 0 | 22 | 22 | 2 | 6 | 9 |
| y | 22 | 0 | 2 | 22 | 26 | 29 |
| h | 22 | 2 | 0 | 22 | 26 | 29 |
| g | 2 | 22 | 22 | 0 | 6 | 9 |
| A | 6 | 25 | 24 | 1 | 0 | 14 |
| S | 9 | 29 | 29 | 9 | 12 | 0 |

N 21

| To | From | | | | | |
|----|------|----|----|---|----|----|
| | z | y | h | g | A | S |
| z | 0 | 1 | 6 | 5 | 4 | 27 |
| y | 1 | 0 | 5 | 6 | 1 | 28 |
| h | 6 | 5 | 0 | 1 | 2 | 14 |
| g | 5 | 6 | 1 | 0 | 3 | 13 |
| A | 4 | 21 | 26 | 5 | 0 | 27 |
| S | 8 | 28 | 28 | 8 | 17 | 0 |

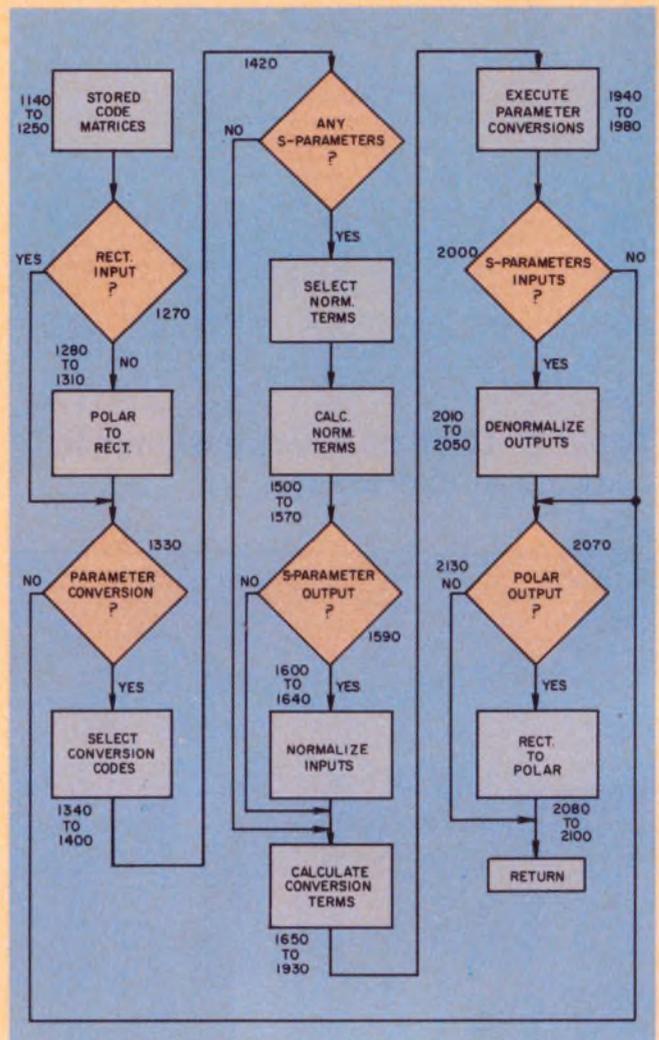
N 22

Denominator Array

| To | From | | | | | |
|----|------|----|----|----|----|----|
| | z | y | h | g | A | S |
| z | 0 | 5 | 4 | 1 | 2 | 14 |
| y | 5 | 0 | 1 | 4 | 3 | 13 |
| h | 4 | 1 | 0 | 5 | 4 | 27 |
| g | 1 | 4 | 5 | 0 | 1 | 28 |
| A | 2 | 2 | 2 | 2 | 0 | 9 |
| S | 13 | 13 | 13 | 13 | 15 | 0 |

D

Note: All numbers refer to results stored in array X.



2. The subroutine AZCON performs the conversions with a coding technique that minimizes redundant calculations.

Table 3. Normalization codes and addresses

| Code No. | Computation | Code No. | Computation |
|----------|-------------|----------|--------------------|
| 0 | None | 4 | $\sqrt{R01 R02}$ |
| 1 | 1 | 5 | $1/R01$ |
| 2 | R01 | 6 | $1/R02$ |
| 3 | R02 | 7 | $1/\sqrt{R01 R02}$ |

| Element | Form of input or output | | | | | |
|----------|-------------------------|---|---|---|---|---|
| | z | y | h | g | A | S |
| x_{11} | 5 | 2 | 5 | 2 | 1 | 0 |
| x_{21} | 7 | 4 | 1 | 1 | 4 | 0 |
| x_{12} | 7 | 4 | 1 | 1 | 7 | 0 |
| x_{22} | 6 | 3 | 3 | 6 | 1 | 0 |

NORM

Note: The computations are stored in the one-dimensional array N by code number; the index of code numbers is stored in the array NORM.

saved, the subroutine may be called by any program as follows:

CALL AZCON (R, P, IC, OC, IF, OF, R01, R02)
where

R = Rectangular input/output, a complex 2×2 matrix (input in rectangular or polar; output in rectangular).

P = Polar form output, a complex 2×2 matrix (optional for OF = 2).

IC = Input Code } 1 = z, 2 = y, 3 = h,
OC = Output code } 4 = g, 5 = ABCD, 6 = S.

IF = Input form } 1 = Rectangular,
OF = Output form } 2 = Polar.

Fortran program for two-port parameter conversion

Mainline program for data input and control

```

1005NDW
110C***** A18: THIS EXAMPLE PROGRAM CONVERTS Z,Y,H,G,ABCD & S-
120C PARAMETERS INTO ANY OF THE ABOVE PARAMETERS USING THE SUB-
130C ROUTINE "AZCON". INPUT MAY BE IN RECTANGULAR OR POLAR FORM,
140C OUTPUT IS GIVEN IN BOTH FORMS. FOR CONVERSIONS TO/FROM S-
150C PARAMETERS, THE NORMALIZING RESISTANCES (R01 & R02) MUST
160C BE SPECIFIED. *****
170C-----
180 COMPLEX R(2,2),P(2,2)
190 INTEGER IC,OC,IF,OF,V(24)
200C-----VARIABLE LIST MATRIX FOR PRINTOUTS-----
210 DATA V /3HZ11,3HZ12,3HZ21,3HZ22,3HY11,3HY12,3HY21,3HY22,
2208 3HH11,3HH12,3HH21,3HH22,3HG11,3HG21,3HG12,3HG22,
2308 3H A,3H C,3H B,3H D,3HS11,3HS12,3HS21,3HS22/
240C-----INITIALIZE MATRICES R & P TO ZERO-----
250 DO 1 J=1,2
260 DO 1 I=1,2
270 1 R(I,J)=P(I,J)=(0.,0.)
280C-----START OF MAIN PROGRAM LOOP-----
290 2 CONTINUE
300C-----READ CONVERSION INSTRUCTIONS-----
310 R01=R02=50
320 PRINT," INPUT,OUTPUT (1=Z,2=Y,3=H,4=G,5=ABCD,6=S)",^*
330 HEAD,IC,OC
340 PRINT," INPUT FORM(1=RECT,2=POLAR)",^*
350 HEAD,IF
360C-----READ R01 & R02 IF ANY S-PARAMETERS-----
370 IF(IC.EQ.6.OR.OC.EQ.6) PRINT," R01,R02=? (DEFAULT=50)",^*
380 IF(IC.EQ.6.OR.OC.EQ.6) READ,R01,R02
390C-----SELECT CORRECT VARIABLES FOR INPUT HEADING-----
400 I1=(IC*4)-3
410 I1=V(I1)
420 I2=V(I1+1)
430 I3=V(I1+2)
440 I4=V(I1+3)
450C-----HEAD INPUT SMALL-SIGNAL PARAMETERS-----
460 PRINT 4,I1,I2,I3,I4
470 4 FORMAT(3X,A3,5X,A3,5X,A2,5X,A3)
480 READ,R
490C-----CALL SUBROUTINE AZCON FOR CONVERSIONS-----
500 CALL AZCON(R,P,IC,OC,IF,2,R01,R02)
510C-----PRINT OUTPUT HEADING-----
520 PRINT 5
530 5 FORMAT(14X,"REAL IMAGINARY",11X,"MAGNITUDE ANGLE")
540C-----SELECT AND PRINT CORRECT OUTPUT VARIABLES-----
550 K=0
560 DO 7 J=1,2
570 DO 7 I=1,2
580 K=K+1
590 IV=(OC*4)-4+K
600 IV=V(IV)
610 PRINT 3,IV,R(I,J),IV,P(I,J)
620 3 FORMAT(3X,A3," = ",G10.4,1X,G10.4,5X,A3," = ",G10.4,1X,G10.4)
630 7 CONTINUE
640 PRINT 9
650 9 FORMAT(/,60(1H-),/)
660 GO TO 2
670C-----
680 END

```

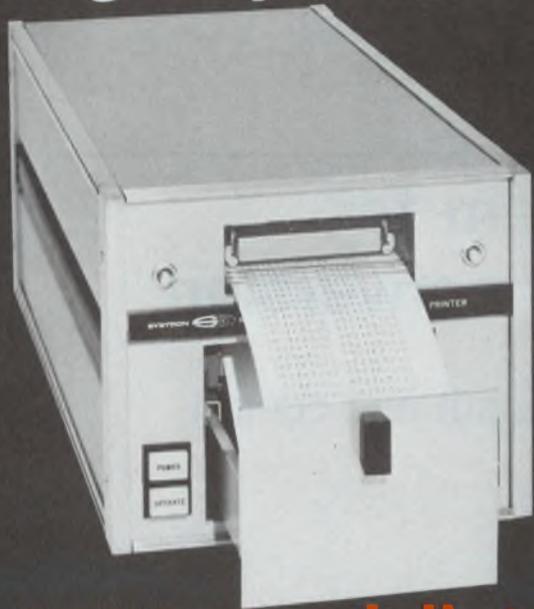
Subroutine for circuit parameter conversion

```

AZCON
1000C*****SUBROUTINE FOR CONVERSION OF Z,Y,H,G,ABCD & S-PARAMETERS*****
1010C R=INPUT/OUTPUT COMPLEX MATRIX
1020C (INPUT IN RECTANGULAR OR POLAR; OUTPUT IN RECT.)
1030C P=POLAR FORM OUTPUT MATRIX (OPTIONAL FOR OF=2)
1040C IC=INPUT CODE; OC=OUTPUT CODE(1=Z,2=Y,3=H,4=G,5=ABCD,6=S)
1050C IF=INPUT FORM; OF=OUTPUT FORM(1=RECT,2=POLAR)
1060C R01 & R02 MUST BE SPECIFIED FOR S-PARAMETER NORMALIZATION
1070C-----
1080 SUBROUTINE AZCON(R,P,IC,OC,IF,OF,R01,R02)
1090 EXTERNAL RECT,POLAR
1100 COMPLEX R(2,2),P(2,2),X(30),DET,RECT,POLAR
1110 INTEGER N1(36),N2(36),N12(36),N21(36),D(36),NORM(24)
1120 INTEGER IC,OC,IF,OF,N1,N2,N3,N4,DD,NM1,NM2,NM3,NM4
1130 REAL N(7)
1140C-----CONVERSION & NORMALIZATION CODE MATRICES-----
1150 DATA N1 /0,4,5,6,1,2R,4,0,6,5,4,27,5,6,0,4,3,1,3,
1160R6,5,4,0,2,14,1,24,25,6,0,2R,7,2,7,27,16,0/
1170 DATA N2 /0,22,22,2,6,9,22,0,2,22,26,29,22,2,0,22,26,29,
1180R2,22,22,0,6,9,6,25,24,1,0,14,9,29,29,9,12,0/
1190 DATA N12 /0,25,5,23,5,10,23,0,23,3,25,30,3,23,0,2,3,5,10,
1200R2,3,2,23,0,25,30,5,26,21,4,0,13,10,30,10,30,11,0/
1210 DATA N22 /0,1,6,5,4,27,1,0,5,6,1,28,6,5,0,1,2,14,
1220R5,6,1,0,3,1,3,4,21,26,5,0,27,8,2R,2R,8,17,0/
1230 DATA D /0,5,4,1,2,14,5,0,1,4,3,13,4,1,0,5,4,27,
1240R1,4,5,0,1,28,2,2,2,2,0,9,1,3,13,1,3,1,15,0/
1250 DATA NORM /5,2,5,2,1,0,7,4,1,1,4,0,7,4,1,1,7,0,6,3,3,6,1,7/
1260C-----CHECK FOR RECTANGULAR INPUT-----
1270 IF(IF.EQ.1) GO TO 1
1280C-----CONVERT INPUT POLAR TO RECTANGULAR-----
1290 DO 2 I=1,2
1300 DO 2 J=1,2
1310 2 R(I,J)=RECT(R(I,J))
1320C-----CHECK FOR PARAMETER CONVERSION-----
1330 1 IF(IC.EQ.OC) GO TO 3
1340C-----SELECT APPROPRIATE CONVERSION CODES-----
1350 K=(OC*6)-4+IC
1360 N1=N11(K)
1370 N2=N21(K)
1380 N3=N12(K)
1390 N4=N22(K)
1400 DD=D(K)
1410C-----CHECK FOR ANY S-PARAMETERS-----
1420 IF(IC.NE.6.AND.OC.NE.6) GO TO 4
1430C-----SELECT APPROPRIATE NORMALIZATION CODES-----
1440 IF(IC.EQ.6) IK=OC
1450 IF(OC.EQ.6) IK=IC
1460 NM1=NORM(IK)
1470 NM2=NORM(IK+6)
1480 NM3=NORM(IK+12)
1490 NM4=NORM(IK+18)
1500C-----CALCULATE NORMALIZATION TERMS-----
1510 N(1)=1
1520 N(2)=R01
1530 N(3)=R02
1540 N(4)=SQRT(R01*R02)
1550 N(5)=1/R01
1560 N(6)=1/R02
1570 N(7)=1/N(4)
1580C-----CHECK FOR S-PARAMETER OUTPUT-----
1590 IF(OC.NE.6) GO TO 4
1600C-----NORMALIZE INPUTS-----
1610 R(1,1)=R(1,1)*N(NM1)
1620 R(2,1)=R(2,1)*N(NM2)
1630 R(1,2)=R(1,2)*N(NM3)
1640 R(2,2)=R(2,2)*N(NM4)
1650C-----CALCULATE CONVERSION TERMS-----
1660 4 DET=R(1,1)*R(2,2)-R(2,1)*R(1,2)
1670 X(1)=R(1,1)
1680 X(2)=R(2,1)
1690 X(3)=R(1,2)
1700 X(4)=R(2,2)
1710 X(5)=DET
1720 X(6)=1.,0.
1730 X(7)=DET*X(1)-X(4)-X(6)
1740 X(8)=DET-X(1)*X(4)-X(6)
1750 X(9)=2*X(2)
1760 X(10)=2*X(3)
1770 X(11)=2*DET
1780 X(12)=2*X(6)
1790 X(13)=DET+X(1)*X(4)+X(6)
1800 X(14)=DET-X(1)-X(4)+X(6)
1810 X(15)=X(1)+X(3)+X(2)+X(4)
1820 X(16)=X(1)+X(3)-X(2)-X(4)
1830 X(17)=-X(1)+X(3)-X(2)+X(4)
1840 X(21)=-X(1)
1850 X(22)=-X(2)
1860 X(23)=-X(3)

```

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```

1870 X(24)=-X(4)
1880 X(25)=-X(5)
1890 X(26)=-X(6)
1900 X(27)=-X(7)
1910 X(28)=-X(8)
1920 X(29)=-X(9)
1930 X(30)=-X(10)
1940-----EXECUTE PARAMETER CONVERSIONS-----
1950 R(1,1)=X(N1)/X(DD)
1960 R(2,1)=X(N2)/X(DD)
1970 R(1,2)=X(N3)/X(DD)
1980 R(2,2)=X(N4)/X(DD)
1990C-----CHECK FOR S-PARAMETER INPUT-----
2000 IF(IC.NE.6) GO TO 3
2010C-----DENORMALIZE OUTPUTS-----
2020 R(1,1)=R(1,1)/N(NM1)
2030 R(2,1)=R(2,1)/N(NM2)
2040 R(1,2)=R(1,2)/N(NM3)
2050 R(2,2)=R(2,2)/N(NM4)
2060C-----CHECK FOR POLAR OUTPUT-----
2070 3 IF(OF.EQ.1) GO TO 999
2080C-----CONVERT OUTPUT RECT. TO POLAR-----
2090 DO 5 I=1,2
2100 DO 5 J=1,2
2110 5 P(I,J)=POLAR(R(I,J))
2120C-----
2130 999 CONTINUE
2140 RETURN
2150 END
2160C*****

```

Auxiliary subroutines for coordinate conversion

RECT

```

10000C*****
10010 COMPLEX FUNCTION RECT(Z)
10020 COMPLEX J,Z
10030 J=(0.0,1.0)
10040 R=Z
10050 A=AIMAG(Z)
10060 A=FP1(A)/180
10070 RECT=R*(COS(A)+J*SIN(A))
10080 RETURN
10090 END
10100C*****

```

POLAR

```

20000C*****
20010 COMPLEX FUNCTION POLAR(Z)
20020 COMPLEX J,Z
20030 J=(0.0,1.0)
20040 R=CABS(Z)
20050 A=AIMAG(Z)
20060 B=REAL(Z)
20070 C=(0.0,0.)
20080 IF(A.EQ.0.AND.B.EQ.0) GO TO 1
20090 C=ATAN2(A,B)
20100 C=C+180/FP1(1.)
20110 ICONTINUE
20120 POLAR=R+J*C
20130 RETURN
20140 END
20150C*****

```

R01 = Input } Impedance-level normaliza-
R02 = Output } tion/denormalization (re-
required for S-parameters
only).

The organization of AZCON is such that very little computation occurs if the user specifies that both input and output are the same two-port parameters (Fig. 2). Otherwise the program performs normalization, computes elements of array X, then performs four computations of the type shown in Eq. 2. The remainder of the calculations involve normalization and polar conversion. The mainline sets the variable OF to a value of two, so the polar-form output is provided. Both rectangular and polar forms of the output are displayed simultaneously. But the user can choose to have the value for OF set to one (rectangular) or two (polar) as needed. ■■

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Analyze TV noise performance by correlating video signal-to-noise ratio with rf noise figure. Then use a spectrum analyzer to perform the measurements.

Noise in TV systems, a source of irritation to the viewer, can be analyzed rapidly by correlation of the signal-to-noise (S/N) ratio in video circuits and the noise figure for rf devices. Rf signal level, modulation degree and detector performance are just a few of the parameters involved in developing the key correlation equation. Calculations using this correlation equation closely match results measured with video noise level meters or spectrum analyzers.

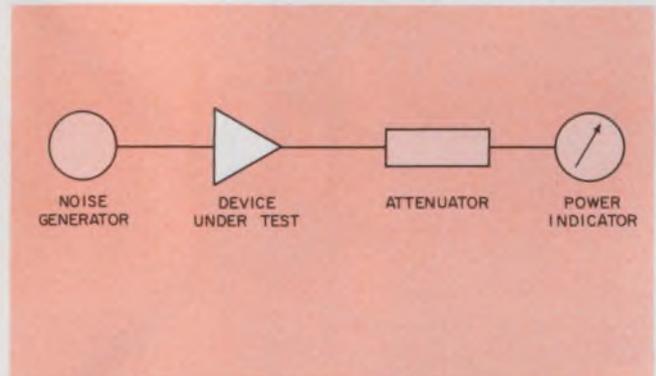
Spectrum analyzers simplify noise measurements, since they include calibrated displays for sync and noise readings. A wideband low-noise preamp is included if the inherent noise of the spectrum analyzer is too high; a selective rejection filter can be added if the video carrier overloads the analyzer.

Noise terms can be misleading

For most transmission systems, the definition of the S/N ratio is rather straightforward; both terms are expressed in rms values. However, for TV video signals, the S/N ratio is defined as the ratio in decibels of the peak-to-peak amplitude of the monochrome picture signal to the rms amplitude of the noise (within the range of 10 kHz to the nominal upper limit of the video frequency band, f_c).

For monochrome TV, the noise band may be shaped by a weighting filter that attenuates the high end of the video band. The theoretical weighting (dB) for "white noise" is 8.5 dB for TV systems B, D, G and H, 6.1 dB for system M and 6.5 dB for system I.¹ TV system M is used in North America, South America and Japan; system I in England, and systems B, D, G and H in continental Europe.

The S/N ratio is a critical factor for modulators, converters and amplifiers designed for monochrome and color TV systems. However, the prescribed weighting filter is not suited for color TV, because the chroma information (being in



1. Noise figure is measured with a calibrated noise generator (NG), adjusted to deliver the same noise power as the device under test (DUT).

the attenuated region) is also sensitive to noise.

Several solutions to this problem have been proposed, but an international standard has not yet been established. In the meantime it has become practice to use just the unweighted S/N ratio. It is generally conceded that 40-dB rms unweighted S/N produces visible impairment on a TV screen, and that better than 60-dB S/N is required for test measurements.

Periodic and impulsive noise is often defined as S/N, except that the noise level is expressed in peak-to-peak value rather than rms. This is called for in CCIR (International Radio Consultative Committee) Rec. 421-2. For example, a single-frequency noise (sine wave) peak-to-peak value is $2\sqrt{2}$ times (~ 9 dB) lower than the rms value. Although the CCIR recommendation restricts the peak-to-peak term to periodic and impulsive noise only, it can also be used for continuous random noise. Thus a specific device may be characterized by different signal-to-noise ratios with up to 17.5-dB difference. For example:

Weighted rms = 58.5 dB,

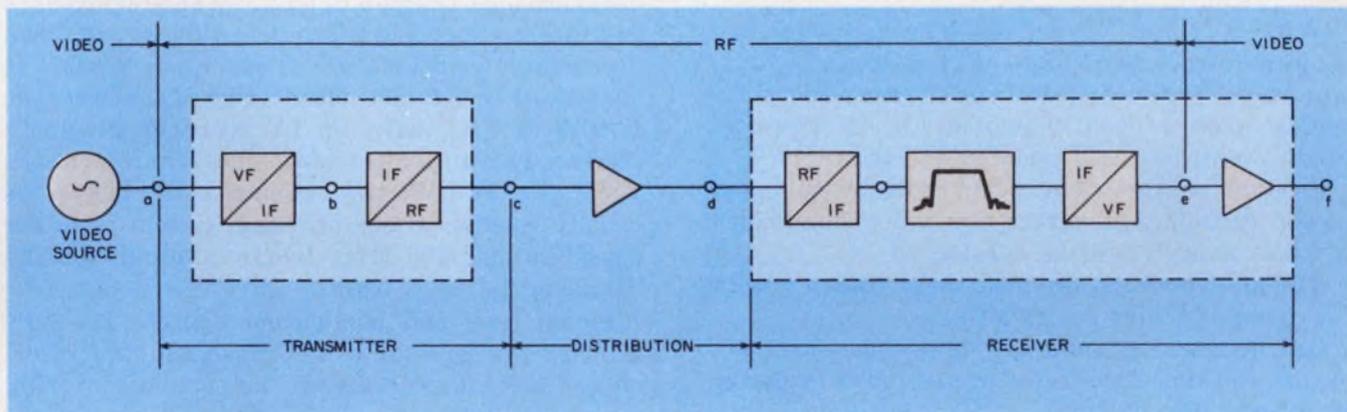
Unweighted rms = 50 dB,

Unweighted peak-to-peak = 41 dB.

Thus whenever a TV signal-to-noise ratio is specified, it is important to state which definition has been used. In the following discussion the unweighted rms signal-to-noise ratio is used.

Noise Figure. Absolute levels of noise power

Jens Simonsen, Development Group Leader, Professional TV Equipment, Philips Radio, Copenhagen, Denmark



2. In a typical TV system, noise figure is used to characterize performance in the rf section while S/N is used

in the video section. But rf noise figures can be transformed to equivalent video S/N ratios.

are usually specified in terms of thermal noise:

$$P_N = kTB, \quad (1)$$

where P_N = Noise power in watts,

k = Boltzman's constant ($1.38 \cdot 10^{-23}$ Joule/K),

T = Temperature in degree Kelvin,

B = Noise power bandwidth

(NPBW) in Hz.

Noise Figure (NF) is a figure of merit relating the actual noise power of a device (referenced to its input terminals) to an equivalent ideal device when the input termination is at the standard reference temperature T_0 (290 K). If the noise power available at the output of an amplifier is called P_N and the power gain G ,

$$NF = \frac{P_N}{kT_0 B G} \quad (2)$$

and, in dB

$$F = 10 \log NF. \quad (3)$$

Expressed in dBm, kT_0 is equal to

$$-174 \text{ dBm/Hz.}$$

When noise is considered in terms of power only and correct matching is assumed, the characteristic impedance is not involved in the equations. However, since TV signals are often given in voltage, dBmV or dB μ V, it is useful to know that

$$0 \text{ dBm} \cong 47 \text{ dBmV} = 107 \text{ dB}\mu\text{V at } 50 \Omega, \quad (4)$$

$$\text{and } 0 \text{ dBm} = 49 \text{ dBmV} = 109 \text{ dB}\mu\text{V at } 75 \Omega. \quad (5)$$

Calculations become simple if dBm and dB terms

are used exclusively.

Suppose we want to determine the noise output power of an amplifier with

$$F = 14 \text{ dB,}$$

$$G = 30 \text{ dB,}$$

$$(\text{NPBW}) = 1 \text{ MHz.}$$

NPBW is 10^6 times 1 Hz, or $10 \log 10^6 = 60$ dB. Thus $P_N = -174 + 14 + 30 + 60 = -70$ dBm.

When two amplifiers are cascaded, the total noise figure is

$$NF_T = NF_1 + \frac{NF_2 - 1}{G_1}. \quad (6)$$

The gain may be greater or less than unity when passive components, lossy cable, attenuators or mixers are included. The noise figure of an attenuator is considered to be equal to its power damping factor. For instance, a 6-dB attenuator has $G = 0.25$ and $NF = 4$.

Measure noise figure simply and reliably

The noise figure is usually measured with a calibrated noise generator, adjusted to deliver the same noise power as the device under test. At this point, the output noise power is doubled (a 3-dB increase).

First, the relative noise P_N is noted with the noise generator switched off (Fig. 1). Next, 3-dB attenuation is added, and the generator is

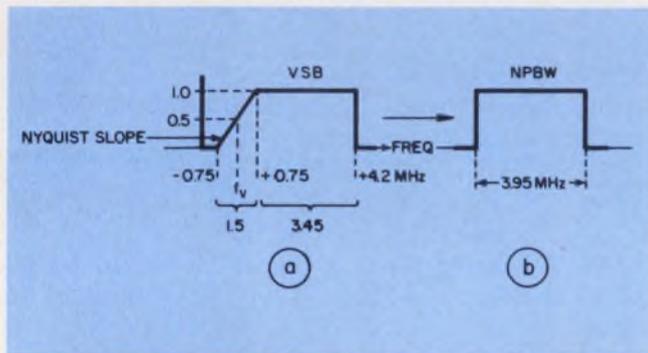
adjusted until the output is at its original level P_N . Noise figure can now be read directly from the scale of the noise generator. The measurement is simple and reliable.

Note that gain and bandwidth need not be known and that the power meter is used only as a reference indicator. The frequency range of the noise generator and the power indicator must exceed that of the device under test (DUT). The noise figure can also be measured by other instruments, such as a spectrum analyzer.

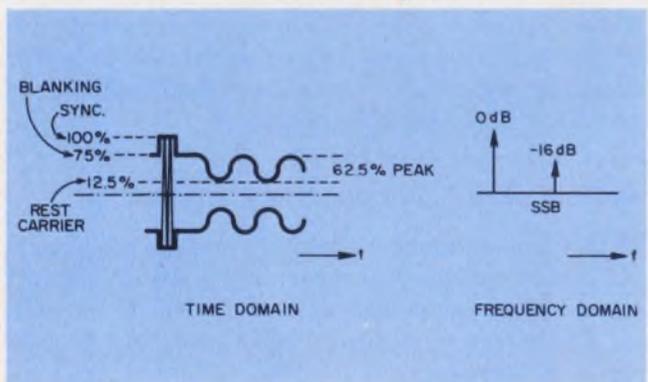
Correlating S/N and noise figure

With S/N and NF defined, let's find a method to transform any rf noise figure to an equivalent video S/N ratio, and vice versa, for a TV system such as that shown in Fig. 2. The correlation will be based upon these conditions:

1. Only continuous random noise with equal power distribution within the video bandwidth ("white noise") will be considered.
2. The system is assumed to be linear within the nominal video bandwidth.
3. The various units are impedance-matched, or else the reflection losses are included in power gain and noise figure.



3. The vestigial sideband (VSB) signal (a) is shaped by a Nyquist filter. The equivalent noise-power-bandwidth (NPBW) is shown (b).



4. The active video information level of a TV rf signal is 62.5% of the total peak-to-peak composite signal, or 16 dB below the carrier.

In practice, these conditions are not critical, since slight deviations will have negligible influence on the result. The derivation is based upon the TV system used in the U.S., system M. Other systems give slightly different figures, as shown in Tables 1 and 2.

Several parameters are involved in the correlation analysis. For example: NPBW due to vestigial sideband (VSB) shaping, rf signal level, modulation degree and detector performance. Let's consider each.

Noise-power-bandwidth (NPBW) and VSB filter shape. The signal-to-noise ratio at rf frequencies must be equal to what we should measure at video frequencies after detection by an ideal measuring demodulator. The VSB signal is shaped in a Nyquist filter (Fig. 3) where the NPBW is 3.95 MHz. A TV test (or Nyquist) receiver has an almost ideal filter characteristic, while the NPBW of a commercial receiver is usually smaller. Since an ideal system is under consideration, 3.95 MHz (corresponding to 65.97 dB above 1 Hz) is used.

Signal level and modulation degree. To characterize the level of a TV rf signal, the reference is normally considered the rms value during sync peaks (Fig. 4). The active video information is only 62.5% when the rest-carrier and sync are subtracted. This corresponds to a sideband that is 16 dB below the sync level. The noise is measured relative to -16 dB, which represents the video information at maximum modulation. According to the CCIR recommendation, the S/N figure should include an artificial reference that is 9 dB higher, because of the ratio of $2\sqrt{2}$ between the rms and the peak-to-peak value of the signal information.

Detector performance. Although a linear system is assumed, obviously the diode detector produces nonlinear distortion of video signals. However, the noise level is generally small compared with the vision carrier, and thus there is no need for a detector correction factor.

Deriving the correlation formula

Let's consider the various elements of the video signal and the noise signal. Fig. 5 displays a graphic representation of the noise and signal composition. Signal and noise must be referenced to the same terminals—in the case of a two-port arrangement, either to the input or to the output. S/N and sync peak level are usually referenced to output, whereas noise figure is referenced to input; so to combine them we must use the power gain factor "G." The thermal noise reference ($1kT_0 = 174$ dBm) and the various system conditional factors can be merged to a common correctional figure "C," and for a two-port arrangement we get:

$$S/N \text{ (dB)} = P_s \text{ (dBm)} - G \text{ (dB)} - F \text{ (dB)} + C \text{ (dB)} \quad (7)$$

For system M (used in the U.S.), "C" is equal to 100.9 dB, and, as shown in Table 1, it is approximately 100 dB for all TV systems. Thus we can use the approximate expression:

$$S/N_{(dB)} \approx P_{S(dBm)} - G_{(dB)} + F_{(dB)} + 100. \quad (8)$$

This rule of thumb is easy to remember, and the rounding off of the "C" figure is suitable for TV systems B, D, G, H, I and M with a maximum error of 1.1 dB.

The correction factor C is composed of k_1 , k_2 , and k_3 . For TV system M:

k_0 = Thermal noise reference at 220 K expressed in dBm,

$$k_0 = 10 \log (1.38 \times 10^{-23} \times 290 \times 10^3) = -173.977 \text{ dBm},$$

k_1 = Signal level and modulation degree.

The rms voltage of one sideband relative to rms voltage during sync peak is

$$\frac{V_{sb}}{V_s} = \frac{0.625}{2\sqrt{2}} \frac{1}{\sqrt{2}} = \frac{0.625}{4} = 0.15625,$$

$$20 \log 0.15625 = 16.124 \text{ dB} (\sim 16 \text{ dB}).$$

The reference level of video S/N is $2\sqrt{2}$ times higher.

Thus:

$$k_1 = 20 \log \left(\frac{0.625}{\sqrt{2}} \right) = 7.093 \text{ dB},$$

k_2 = NPBW of Nyquist filter.

$$\text{In general, NPBW} = \frac{1}{2\pi} \int_{\omega=0}^{\omega=\infty} |A(\omega)|^2 d\omega,$$

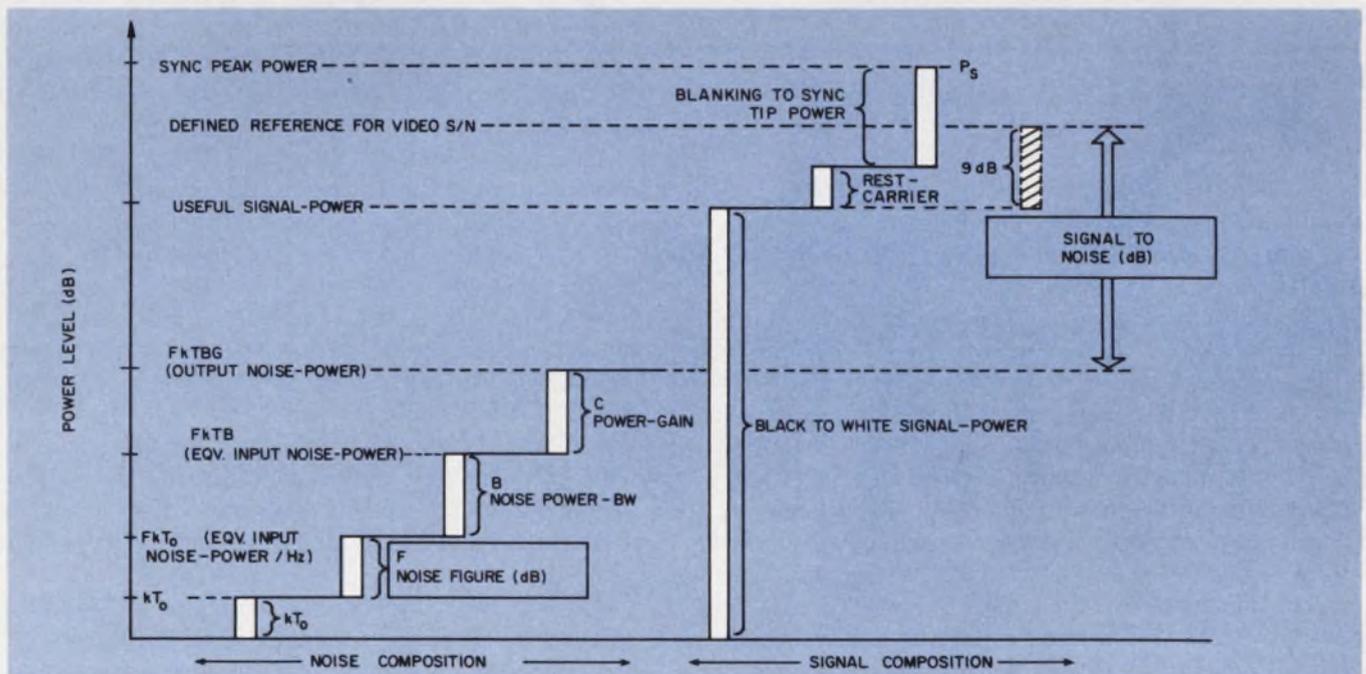
where $A(\omega)$ is the amplitude transfer function, (Fig. 3a),

$$\begin{aligned} \text{NPBW} &= \int_0^{1.5 \text{ MHz}} f^2 df + \int_0^{3.45 \text{ MHz}} 1 df \\ &= 1.5 \times \frac{1}{3} + 3.45 = 3.95 \text{ MHz}, \end{aligned}$$

Table 1. Correction factors for worldwide TV systems

| TV system | White-level† (%) | Blanking-level† (%) | Sideband (%) | (dB) | K1 (dB) | VSB† (MHz) | F(C)† (MHz) | NPBW (MHz) | K2 (dB) | Correction (dB) |
|-----------|------------------|---------------------|--------------|---------|---------|------------|-------------|------------|---------|-----------------|
| B.G | 10 (10-12.5) | 75 (72.5-77.5) | 65.0 | -15.783 | -6.752 | .75 | 5.0 | 4.750 | 66.767 | C = 100.4583 |
| D | 10 (10-12.5) | 75 (72.5-77.5) | 65.0 | -15.783 | -6.752 | .75 | 6.0 | 5.750 | 67.597 | C = 99.6285 |
| I | 20 | 76 | 56.0 | -17.077 | -8.047 | 1.25 | 5.5 | 5.083 | 67.061 | C = 98.8692 |
| I* | 20 | 76 | 56.0 | -17.077 | -8.047 | 1.25 | 5.0* | 4.583 | 66.612 | C = 99.3189 |
| M | 12.5 (10-15) | 75 (72.5-77.5) | 62.5 | -16.124 | -7.093 | .75 | 4.2 | 3.950 | 65.966 | C = 100.9186 |

*) Luminance BW = 5 MHz for S/N measurement for System I acc. to CCIR Rec. 451-1
 †) Data from CCIR rep. 308-2



5. Graphic representation of noise and signal components identifies each element's contribution to the total level.

$$k_2 = 10 \times \log(3.95 \times 10^6) = 65.966 \text{ dB}$$

$$C = k_1 - k_0 - k_2 = 7.093 + 173.977 - 65.966 = 100.918 \text{ dB}$$

Here is an example to illustrate the correlation. A distribution amplifier with 14-dB power gain and 12-dB noise figure has an rms sync peak output of 100 mV into 75 Ω . What is the equivalent signal-to-noise ratio?

$$\text{(From Eq. 5)} \quad P_s = 100 - 109 = -9 \text{ dBm}$$

$$\text{(Using Eq. 8)} \quad S/N \cong -9 - 14 - 12 + 100 = 65 \text{ dB}$$

If the P_s and F terms are referenced to the same port, "G" in Eq. 7 and 8 must be omitted.

As another example, assume the output signal from a TV converter is 100 mV into 50 Ω with $S/N = 60$ dB. The signal is distributed via a lossy coaxial cable to a receiver having an input noise figure of 14 dB (Fig. 6). How much cable attenuation can be tolerated when the signal-to-noise ratio must be 50 dB?

First, calculate noise from the converter:

$$\text{(From Eq. 8)} \quad F_1 = P_1 - S/N_1 + 100 = -7 - 60 + 100 = 33 \text{ dB} \sim 2000 \text{ kT}_0$$

Next, the noise figure at point a:

$$NF_a = NF_1 + \frac{1}{G}$$

The system noise figure is found next:

$$\text{(Using Eq. 8)} \quad F_T = -7 - 50 + 100 = 43 \text{ dB} \sim 20,000 \text{ kT}_0$$

$$\text{(Using Eq. 6)} \quad NF_T = NF_a + \frac{NF_3 - 1}{G}$$

$$20,000 = 2000 + \frac{1}{G} + \frac{25 - 1}{G}$$

$$G = \frac{25}{20,000 - 2000} = \frac{1}{720} \sim -28.6 \text{ dB}$$

With 28.6-dB cable loss, the signal level at point b will be:

$$P_b = -7 - 28.6 = -35.6 \text{ dBm}$$

and (using Eq. 4) $V_b = -35.6 + 47 = 11.4 \text{ dBmV} = 3.7 \text{ mV}$.

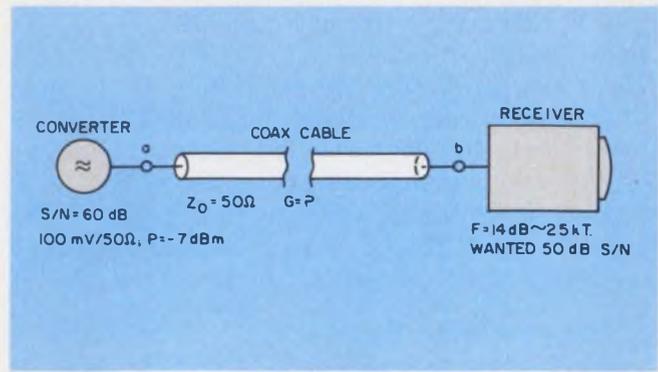
Proceeding to a trial calculation for S/N :

$$\text{(From Eq. 6)} \quad NF_a = 720 + \frac{25 - 1}{1/720} = 18,000 \text{ kT}_0$$

$$NF_T = 18,000 + 2000 = 20,000 \text{ kT}_0 \sim 43 \text{ dB}$$

$$\text{(Eq. 8)} \quad S/N = 35.6 - (-28.6) - 43 + 100 = 50 \text{ dB}$$

Experiments have verified the validity of Eq. 7. In a system like that of Fig. 2, calibrated "white noise" was added to test points a to f. By means of a noise generator, the noise figure was measured at various points, and, after detection, S/N was measured with a video signal-to-noise measuring instrument (video noise level meter). The measured results were in excellent agreement with calculations made with Eq. 7 for small signal-to-noise ratios (deviation of only 0.5 dB for 30 dB S/N). At higher S/N , the error became larger (up to 2 dB at $SN = 60$ dB) due to inherent noise in the test setup. After corrections



6. The acceptable cable attenuation may be found once the converter and receiver noise are calculated.

(using Eq. 6), it was shown that the correlation formula (Eq. 7) could be used to the sensitivity limit of the measuring instruments. The experiments also indicated that a spectrum analyzer could be used for noise measurement with almost the same accuracy as a standard noise generator and a video noise level meter.

Spectrum analyzers for S/N tests

Spectrum analyzers (SPA), widely used for measurement of spurious and intermodulation signals, can also be used for noise measurements. Since modern spectrum analyzers are calibrated in dBm, the sync level, P_s , can be read directly from the screen, and noise power, P_N , can be found from a simple calculation. The measurement technique has been described² with the necessary correction factors for spectrum-analyzer detector performance, logarithmic shaping and NPWB vs i-f, 3 dB BW. For a 3-dB of 10 kHz, these factors are calculated as $k_3 = 38.3$ dB.

As an example, consider a spectrum analyzer connected to an amplifier with a noise level of -70 dBm measured in a 10-kHz i-f bandwidth. To find the noise power per Hz expressed in dBm:

$$-70 \text{ dBm} - 38.3 \text{ dB} (k_3) = -108.3 \text{ dBm/Hz}$$

If the power gain of the amplifier is 20 dB, we can calculate the noise figure as before:

$$F = 174 - 108.3 - 20 = 45.7 \text{ dB}$$

If the spectrum analyzer is always used with the same i-f bandwidth position, the correction figures can easily be included in Eq. 7 for direct calculation of S/N . For 10-kHz i-f bandwidth, the S/N for system M is:

$$S/N_{dB} = P_{s(dBm)} - P_{N(dBm)} - 34.8 \text{ (dB)} \quad (9)$$

when P_N is the indicated noise level in 10-kHz bandwidth and a 100-Hz video filter is used for averaging. The figure may be rounded to 35 dB.

The constant in Eq. 9 is derived as follows: The measurement is made with an HP spectrum analyzer. We must add 2.5 dB for log shaping and detector performance. The NPBW is 1.2

Table 2. Constants for S/N spectrum-analyzer measurements

| TV System | K3 (dB) | C2 (dB) | Video NPBW (MHz) | K4 (dB) | V(REF) (V) | K5 (dBm) | C3 (dB) |
|-----------|---------|---------|------------------|---------|------------|----------|---------|
| B-G | -38.3 | 35.219 | 5.0 | 66.990 | .700 | 8.151 | 20.538 |
| D | -38.3 | 36.049 | 6.0 | 67.782 | .700 | 8.151 | 21.330 |
| I | -38.3 | 36.808 | 5.5 | 67.404 | .700 | 8.151 | 20.952 |
| I* | -38.3 | 36.358 | 5.0* | 66.990 | .700 | 8.151 | 20.538 |
| M | -38.3 | 34.759 | 4.2 | 66.232 | .714 | 8.323 | 19.609 |

*) Luminance BW = 5 MHz for S/N measurements for System I acc. to CCIR Rec. 451-1

times the 3-dB i-f bandwidth.

$$k_3 = 2.5 \text{ dB} + 10 \log \left(\frac{1 \text{ Hz}}{1.2 \times \text{BW}_{3\text{dB}}} \right).$$

Therefore P_N measured in 10-kHz i-f bandwidth converted into 1-Hz bandwidth is:

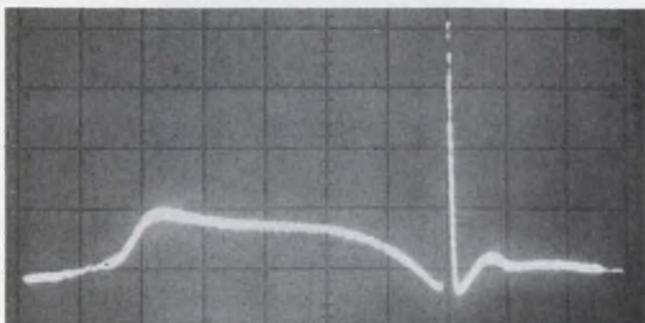
$$k_3 = 2.5 \text{ dB} + 10 \log \left(\frac{1}{10^4 \times 1.2} \right) = -38.3 \text{ dB}$$

$$C_2 = k_2 + k_3 - k_1 \\ = 65.966 - 38.3 + 7.093 = 34.759 \text{ dB.}$$

Now let P_s be +5 dBm. Then $\text{SN} = +5 - (-70) - 35 = 40 \text{ dB}$.

With high S/N ratios, the inherent noise of the spectrum analyzer may be too large, but a wideband low-noise preamplifier can be used to improve the sensitivity. Another problem often met is overloading of the spectrum analyzer input caused by the video carrier. For linear devices, the carrier can simply be switched off during measurement of P_N . This method is not applicable to nonlinear devices because the noise level may change when the carrier is removed. In this case a rejection filter tuned to the carrier frequency can be used before the spectrum analyzer to prevent overloading (Fig. 7).

If resistive pads are used to match a 75- Ω system to a 50- Ω spectrum analyzer, the power damping factor must be included. Power attenuation is different from voltage damping; power damping = 5.7 dB for 75 Ω to 50 Ω minimum loss-pad.³ However, it is generally unneces-



7. A notch filter attenuates the video carrier to avoid spectrum-analyzer overload during noise measurement.

sary to use such matching pads during the measurement of noise level, since a VSWR of 1.5 (50 Ω to 75 Ω) produces only a power reflection loss of 4% (-0.18 dB). The 5.7-dB higher sensitivity in the unmatched case is often preferred and justifies the 0.18-dB error.

It is also interesting to note that the spectrum analyzer can be used for direct S/N measurements of video signals. Here we obtain another correction figure, because the VSB filter and the modulation degree are not involved. In addition the monochrome signal is assumed to be the standard level of 0.714 V_{pp} at 75 Ω . So the reference level is 0.714 V rms = +8.3 dBm. If we again standardize on 10-kHz i-f bandwidth for the spectrum analyzer during measurement of P_N , the equation for system M (4.2 MHz NPBW) becomes

$$S/N_{(\text{dB})} = -P_{N(\text{dBm})} - 19.6_{(\text{dB})}. \quad (10)$$

Again, a rounding to -20 dB is appropriate.

The constant for Eq. 10 is derived as follows:

NPBW = 4.2 MHz,

$$k_1 = 10 \times \log (4.2 \times 10^6) \simeq 66.2 \text{ dB,}$$

Video S/N reference = 0.714 V rms in 75,

$$P_r = \frac{.714^2}{75} \times 1000 = 6.797 \text{ mW,}$$

$$k_5 = 10 \log 6.797 = 8.3 \text{ dBm,}$$

$$C_3 = k_3 + k_1 - k_5,$$

$$C_3 = -38.3 + 66.2 - 8.3 = 19.6 \text{ dB.}$$

In Table 2 the correction figures for Eqs. 9 and 10 are shown for other TV systems.

As a final example: From a video amplifier with nominal black/white level = 0.714 V_{pp} , we read $P_N = -65 \text{ dBm}$ in 10-kHz BW from a spectrum analyzer. What is the rms unweighted S/N?

$$S/N = -(-65) - 20 = 45 \text{ dB.} \quad \blacksquare$$

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2. "Measure Oscillator Spectral Purity with a Spectrum Analyzer," *Microwave Journal*, Vol. 16, No. 4, April, 1973, pp. 53-56.
3. Simonsen J., "How to Make Your Own Matching Pads," T & M News Supplement PTV series, Philips, Test and Measuring Dept., 1973.

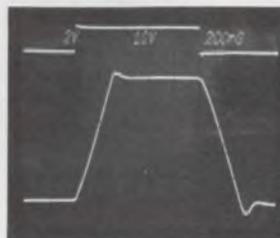
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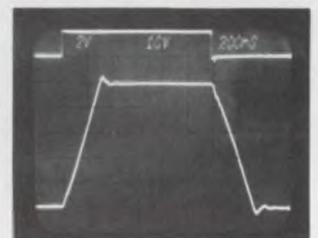
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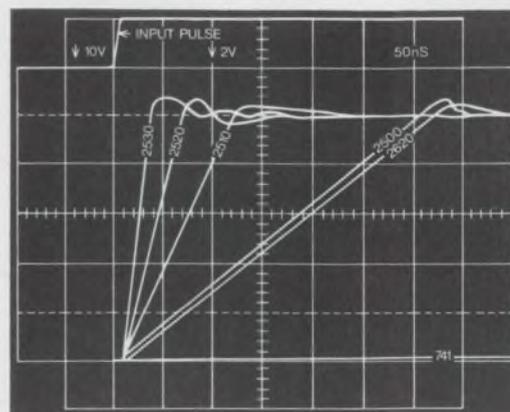
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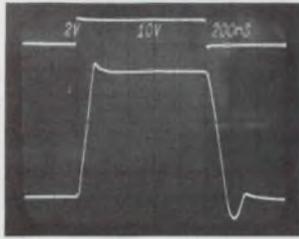
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| Gain Bandwidth Product | 100 | 100 | 100 | 12 | 12 | 12 |
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| Voltage Gain | 100k | 80k | 80k | 20k | 15k | 15k |
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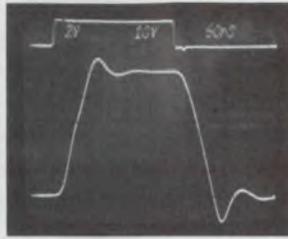
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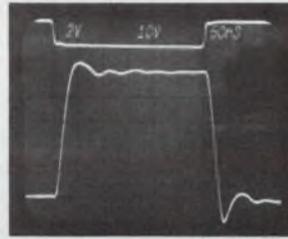
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| ± 50 | ± 40 | ± 40 | ± 100 | ± 80 | ± 80 | ± 280 | ± 250 | V/μS (MIN) | |
| 750 | 600 | 600 | 1500 | 1200 | 1200 | 4000 | 4000 | kHz (MIN) | |
| 12 | 12 | 12 | 20 | 20 | 12 | 70 | 70 | MHz (TYP) | |
| 250 | 250 | 250 | 200 | 200 | 200 | 500 | 500 | ns (TYP) | |
| 10k | 7.5k | 7.5k | 10k | 7.5k | 7.5k | 100k | 100k | V/V (MIN) | |
| 200 | 250 | 250 | 200 | 250 | 250 | 100 | 200 | nA (MAX) | |
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INFORMATION RETRIEVAL NUMBER 52

ECL 10,000 interconnects economically.

Controlled edge speeds and low-voltage swings allow use of conventional boards for this increasingly popular logic.

The ECL 10,000 high-speed logic family doesn't necessarily need expensive packaging. Surprised? For many applications, inexpensive double-sided PC or wire-wrappable boards can be used in place of costly multilayer boards.

Once priced significantly higher than other logic families, ECL 10-k used to be confined to expensive systems. But prices have fallen as a result of multiple sourcing, plastic packaging and volume manufacturing. As a result, logic designers are turning increasingly to the high-speed ECL family. They're using ECL 10-k to upgrade older systems or as an alternative to TTL for new systems.

Earlier, higher-speed, ECL circuits require multilayer boards and stripline wiring to take advantage of the logic speed. ECL 10-k can use the less-expensive boards because of controlled edge speed, input pull-down resistors and voltage reference sources. Features like these simplify interconnect considerations, which conveniently break down into the following:

- Grounding.
- Coupling and loading.
- Terminations.
- Crosstalk.
- System connections.

The most popular bipolar logic lines are ECL 10-k, Schottky-TTL and standard TTL. Table 1 compares the family characteristics that affect wiring and interconnects.

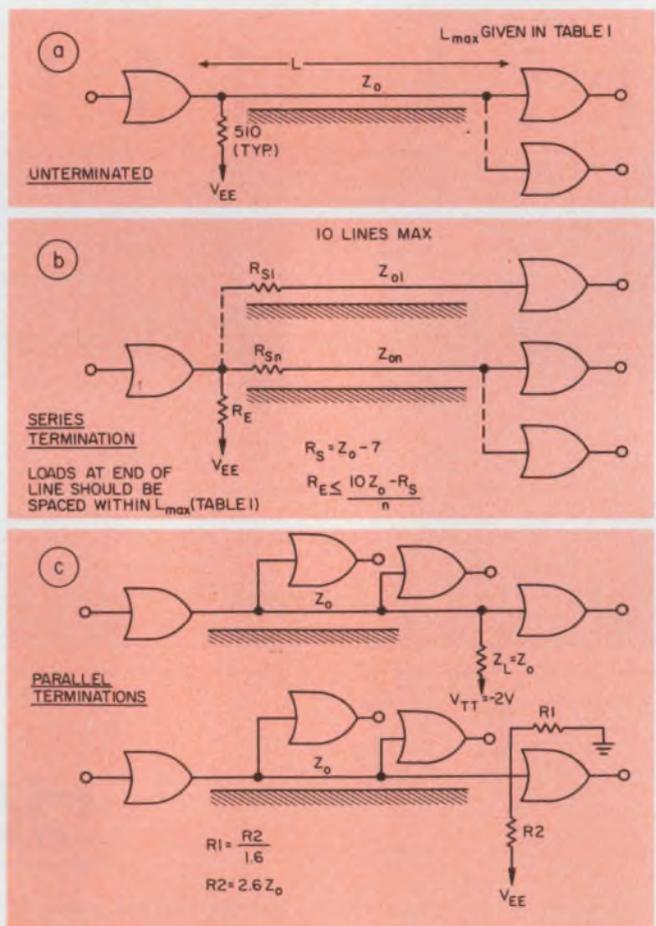
Grounding differs with board types

Current transition (ΔI) during a change in logic level is about the same for all three logic families. ECL 10-k, with its lower noise margin, requires a lower inductance ground connection than the TTL families to maintain comparable performance.

For double-sided PC boards, this means maintaining a ground plane over 60 to 70% of one

side of the board and using a 0.1-in. minimum line width for V_{EE} connections. For wire-wrappable, or solderless, boards or backplanes, the designer can construct a 1-in. ground grid by bussing together ground pins at 1-in. intervals.

In addition to low impedance, a ground plane or grid provides a controlled characteristic impedance for signal lines. At the connector interface to the board, about 10% of the pins should

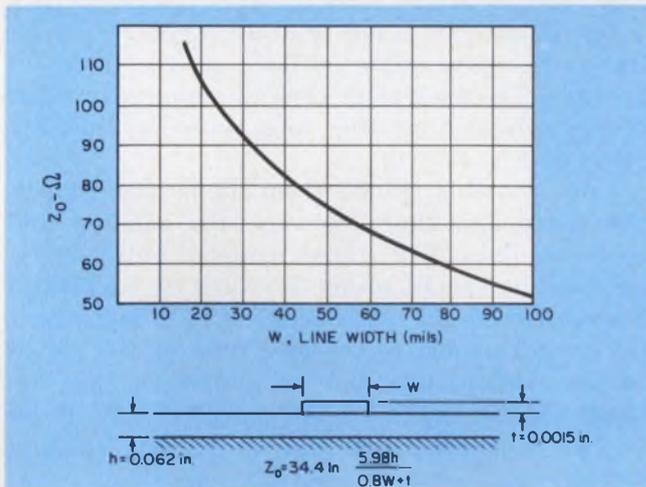


1. Series or parallel terminations on ECL 10-k circuits eliminate noise problems caused by mismatches on unterminated lines (a). In the series configuration (b), each load at the receiving end reduces the noise margin by 0.16 (R_s) millivolts. In the parallel configuration (c), up to 16 distributed loads can be accommodated. Each load adds about 100 ps of delay.

Robert Jaeger, Applications Engineer, Signetics Corp., 811 E. Arques Ave., Sunnyvale, Calif. 94086.

be reserved for grounds.

Decoupling requirements can be met by a $1\text{-}\mu\text{F}$ bypass capacitor on each board at the power-supply inputs. Ground and V_{EE} inputs should be close to each other to minimize lead inductance in the bypass capacitor. Generally every four or five packages on a board require a $0.01\text{-to-}0.1\text{-}\mu\text{F}$ rf capacitor. Also, lines connecting V_{EE} pins on each package to the power supply should be a



2. A line on a PC board produces a controlled value of characteristic impedance (Z_0). The values and expression for Z_0 assume a dielectric constant that is typical of glass-epoxy material.

minimum of 0.1-in. wide.

Inputs to ECL 10-k circuits typically require only $220\ \mu\text{A}$ at the logic ONE level. Hence dc loading doesn't impose restrictions. Instead loading is constrained solely by ac considerations.

Lines don't require terminations

Unlike some of the earlier ECL families, ECL 10-k doesn't necessarily require terminated lines (Fig. 1). However, the ability to drive terminated lines allows a reduction of line noise.

Terminations are not required when the two-way delay of the lines is less than the rise time of the signal edge. Each ECL 10-k input has a 3-pF input capacitance. Since capacitive loading of the line increases the line delay, the maximum allowable line length decreases with increased loading. This capacitive loading has less effect on lines with low characteristic impedance (Z_0). Table 2 shows the maximum line length that should be used for unterminated lines with different values of Z_0 and loading.

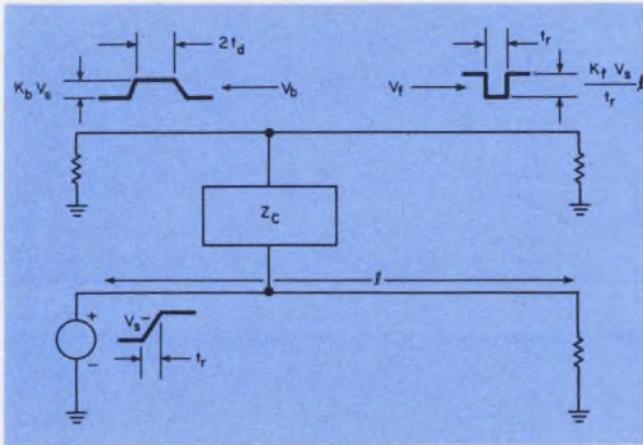
The value of Z_0 for lines on a wire-wrappable board is about $120\ \Omega$. However, this can vary from 100 to 180, depending on distance from the ground plane, proximity of adjacent wires and configuration of the ground grid.

The value of Z_0 for a line on a PC board is much more closely controlled. The impedance de-

Table 1. Comparison of three logic families

| Parameter | TTL | Schottky TTL | ECL 10,000 |
|---|-----|--------------|------------|
| Propagation delay—ns | 10 | 3 | 2 |
| Voltage swing—V | 3.5 | 3.5 | 0.8 |
| Noise margin—V | 0.4 | 0.3 | 0.125 |
| % | 11 | 9 | 15 |
| Rise/fall time—ns | 8 | 2 | 2 |
| Edge speed ($\Delta V/\Delta T$)*—mV/ns | 440 | 1750 | 400 |
| Edge current (ΔI)**—mA | 16 | 20 | 17 |

* affects crosstalk
 ** affects grounding and decoupling



3. Adjacent, parallel lines produce crosstalk noise. Impedance Z_c represents a combined mutual inductance and capacitance. This impedance is responsible for the coupling of signals from one line to another.

depends on board thickness and line width. Fig. 2 shows a plot of Z_0 vs line width for 0.062-in. boards having a ground plane on one side.

'Ringing' can be eliminated

The ability to terminate signal lines with ECL 10-k circuits provides several design benefits. Terminations can be used to eliminate line reflections, or ringing, completely when line lengths exceed the values in Table 2. In addition terminations can reduce crosstalk between parallel signal lines.

Line reflections occur when the impedance of the load at the receiving end of the line (Z_L) differs from the characteristic impedance of the line (Z_0). The value of the reflected voltage (V_r) depends on the impedance mismatch, as follows:

$$V_r = V_{line} \frac{(Z_L - Z_0)}{(Z_L + Z_0)}$$

Table 2. Allowable unterminated line lengths depend on fanout

| $Z_0 - \Omega$ | Maximum line length— L_{MAX} (inches) | | | |
|----------------|---|----------|----------|----------|
| | fanout=1 | fanout=2 | fanout=4 | fanout=8 |
| micro-strip: | | | | |
| 50 | 8.3 | 7.5 | 6.7 | 5.7 |
| 68 | 7.0 | 6.2 | 5.0 | 4.0 |
| 75 | 6.9 | 5.9 | 4.6 | 3.6 |
| 90 | 6.5 | 5.4 | 3.9 | 3.0 |
| 100 | 6.3 | 5.1 | 3.6 | 2.6 |
| back-plane: | | | | |
| 100 | 6.6 | 5.4 | 3.8 | 2.8 |
| 140 | 5.9 | 4.3 | 2.8 | 1.9 |
| 180 | 5.2 | 3.6 | 2.1 | 1.3 |

For any value of Z_L close to Z_0 , the reflection drops to a negligible value; for $Z_L = Z_0$, $V_r = 0$. Also, the polarity of the reflection is the same as the driving signal when $Z_L > Z_0$, and it is reversed when $Z_L < Z_0$.

Both series and parallel terminations can be used in ECL 10-k systems (Fig. 1). In a series termination a resistor is placed in series with the signal line at the driving end. For parallel terminations, a load resistor is used between the line and V_{EE} (or a V_{TT} voltage) at the end of the line.

Crosstalk can be controlled

Crosstalk results from the coupling of a pulse on one signal line to an adjacent line. It can be controlled readily when ECL 10-k is used. The logic family has a linear input impedance, low output impedance and capability for signal-line termination. These features permit crosstalk to be analyzed and minimized—which is why ECL 10-k is called the "quiet" high-speed family.

The simplest way to minimize crosstalk is to run adjacent signal lines at right angles to each other. Where this is not feasible, a general guideline is to space adjacent lines as far apart as possible. The equivalent circuit, representing the coupling from a sending to a receiving line, appears in Fig. 3.

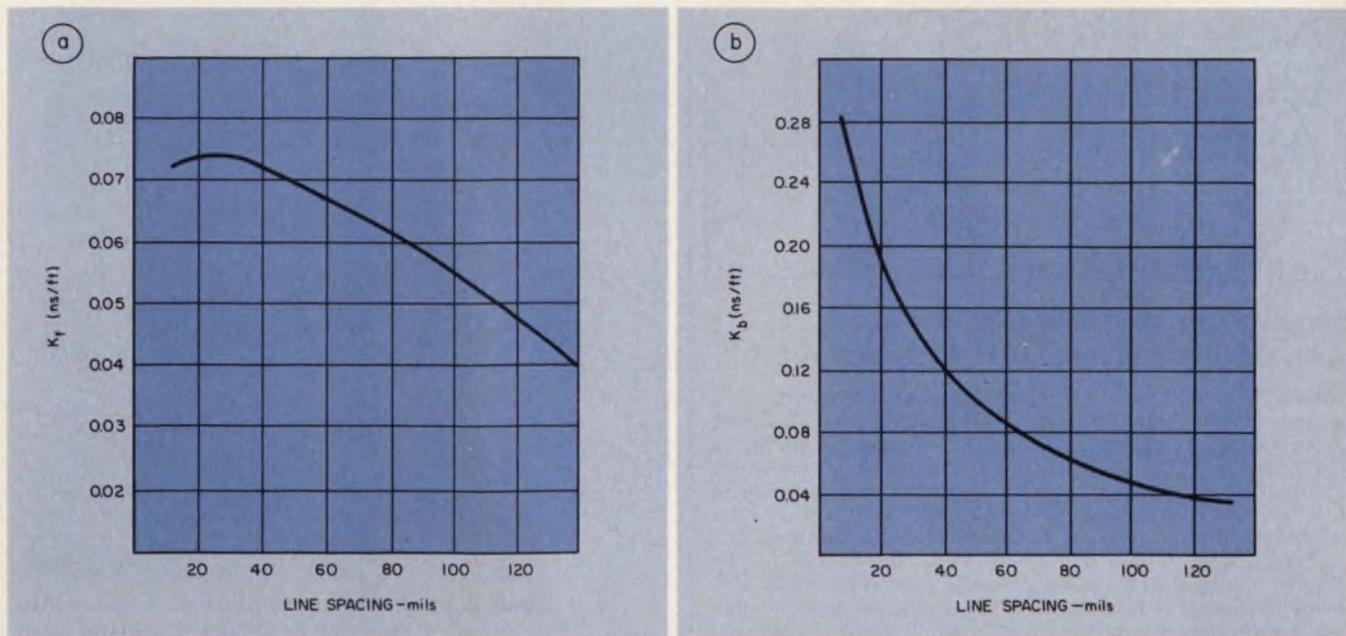
Cross-coupling results from the mutual inductance and capacitance between the sending and receiving lines. The signals coupled onto the receiving line can be either forward or backward. Forward crosstalk (V_f) consists of a pulse having a width equal to the rise time of the signal on the sending line and an amplitude that depends on the length of the parallel lines. Pulse V_f has a polarity opposite to V_s , the sending signal.

Forward crosstalk isn't usually significant in ECL 10-k systems, because of the small $\Delta V/\Delta T$ on the signal edge. As indicated in Table 1, the value of $\Delta V/\Delta T$ in ECL 10-k is about the same as that for TTL and less than a quarter the value for Schottky-TTL.

The backward crosstalk pulse (V_b) has a width that is equal to twice the propagation delay of the sending line. The amplitude depends on the line spacing only and is independent of the length of the parallel lines or signal rise time. The rise and fall times of V_b equal the rise and fall times, respectively, of V_s . Fig. 4 shows a plot of these times vs line spacing for 0.062-in. boards.

Fig. 5 shows the waveforms of the backward crosstalk signal for a variety of terminations on the receiving line. Since V_b always travels in a direction opposite to V_s , a termination at the lefthand side of each diagram completely absorbs

(continued on page 94)



4. Rise and fall times of crosscoupled signals are drawn for the two possible cases: forward crosstalk (a)

and backward (b). The first type has a small $\Delta V/\Delta T$ on the signal edge, making it less significant.

| DRIVE SIGNAL ON SENDING LINE | | |
|------------------------------|-----------------|-----------------|
| CONFIGURATION | WAVEFORM AT X=0 | WAVEFORM AT X=l |
| | | |
| | | |
| | | |
| | | |
| | | |
| | | |

5. Waveforms of backward-crosstalk signals are affected by the terminations used. Regardless of the type of transmission line employed, a termination with an im-

pedance value equal to the line's characteristic impedance completely absorbs any incident reflection. Curves of constants K_b and K_f appear in Fig. 4.

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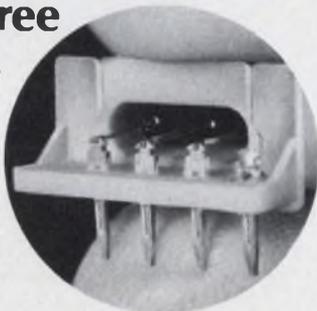
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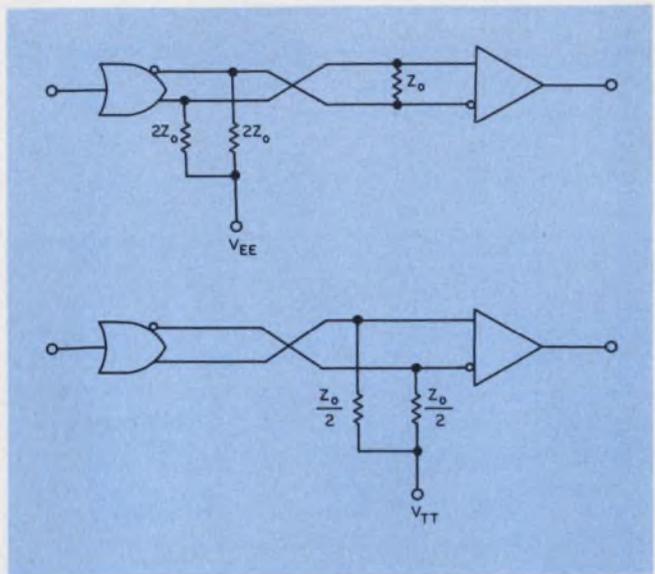
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6. Twisted-pair lines should be terminated in one of two ways. Up to 50 ft of line can be driven at greater than 100 MHz.

any coupled V_b .

Series-terminated lines generate less crosstalk than parallel-terminated ones. The reason is that only half the logic swing is sent down a series-terminated line.

Coax and twisted-pair to connect boards

Generally coax or twisted-pair are used for ECL interconnections between boards. Coax should have a characteristic impedance in the range of 50 to 100 Ω , and each end of transmission-line returns should be grounded. Twisted-pair—AWG24 to 28, twisted 30 turns per foot—yields a satisfactory Z_0 of about 110 Ω .

If ribbon cable is used, select types having low attenuation at the maximum operating frequency. The maximum recommended attenuation for ECL 10-k is 2.5 dB, which may limit the cable length to 15 ft or less. Alternate leads in the ribbon cable should be grounded to maintain the transmission-line effect.

At connectors, use multiple-ground pins, equally spaced, to reduce the connection impedance to negligible amounts.

Very long lines—20 ft or more—can be driven differentially to maintain noise immunity at high frequencies. The complementary outputs of gates can be used for differential drive; and a 10114/15 ECL circuit can be employed as the line receiver.

Fifty feet of twisted pair can be driven differentially at frequencies over 100 MHz. Differentially driven lines should be terminated as shown in Fig. 6, and the fanout should be limited to four. The twisted-pair lines should be shielded when the common-mode noise on the pair exceeds ± 1 V. ■■

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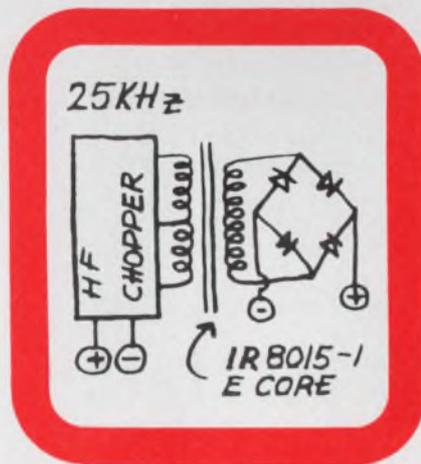
facts you need; component specifications, temperature characteristics and application data. The design procedure we outline in this guide and an afternoon of your time will tell you if higher frequency operation is the solution to your specific power supply requirements.

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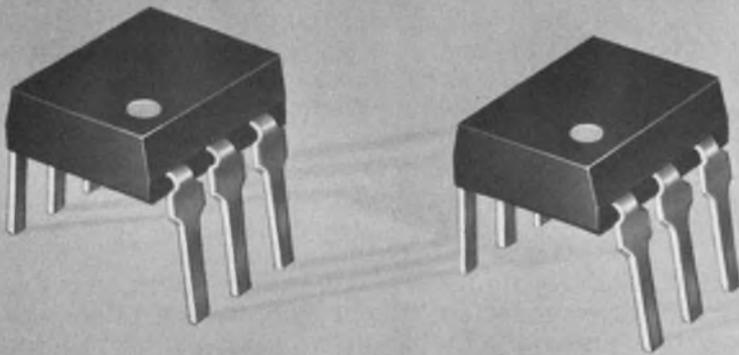


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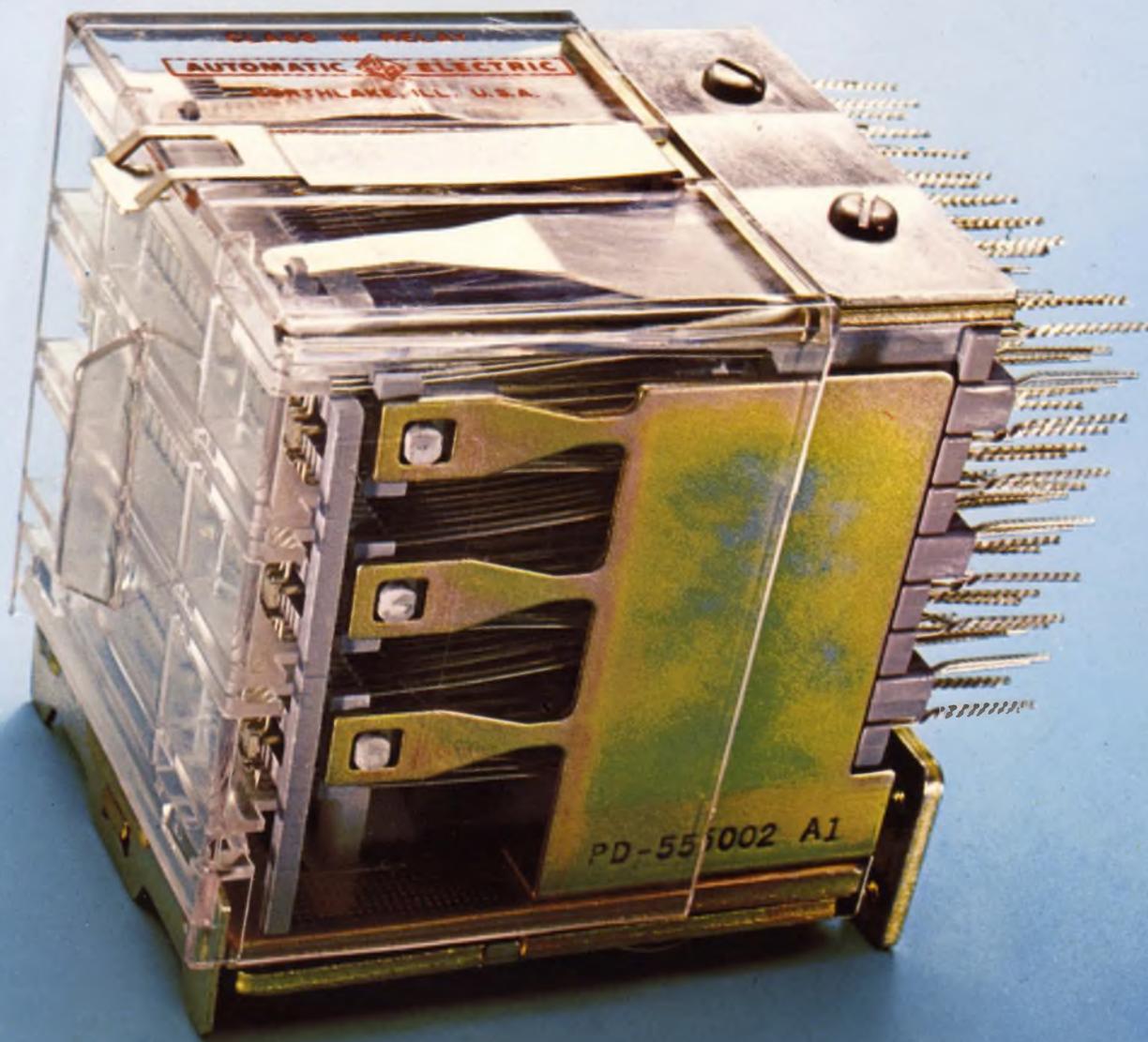
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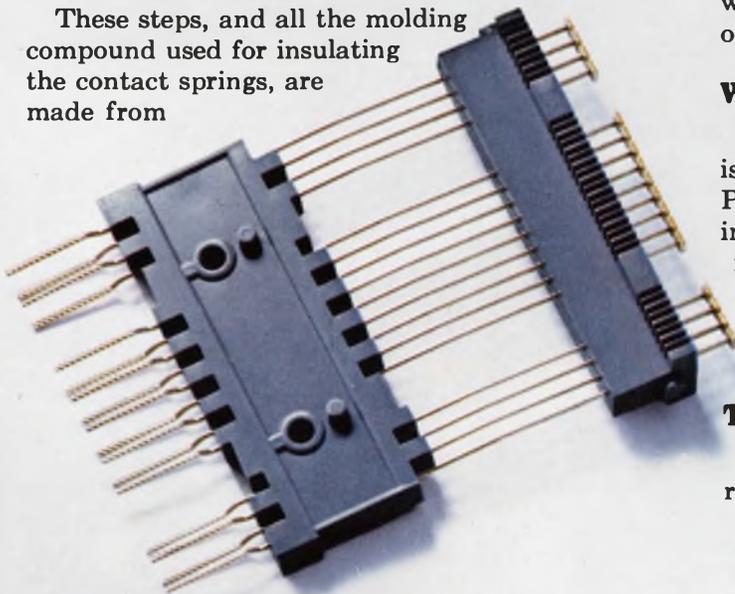
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All those staggered steps you see on the side were put in to raise the breakdown voltage between terminals. These molded steps add extra creepage distance between the terminals. This really counts for high voltage testing, or when using our Class W in unfavorable ambient conditions.

These steps, and all the molding compound used for insulating the contact springs, are made from



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Each of our long wire-spring contacts has an independent twin with the same function. One tiny particle of dust could prevent contact on other relays. Not with our Class W. You can be sure one of the twins will function. That's back-up reliability.

The twin contacts are twisted together at the terminal end. Then we give them a spanking (you might call it swedging) to provide solderless wrap.

We're for independence.

Our springs are longer, because the longer the spring, the more independent they get. And the better contact they make. Don't forget, the wire-spring relay is the most reliable way to get a permissive make or break contact. You can rely on it.

The middle contact springs have to be stationary. To make sure they stay that way forever, we actually mold them between two thick pieces of DAP on both ends. Just try to move one.

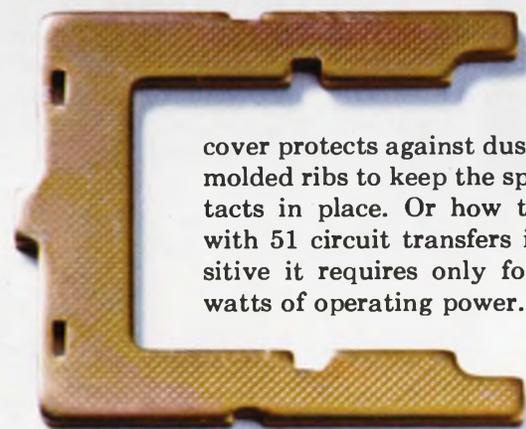
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Look to asynchronous sequential logic

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Before you design your next sequential logic circuit, take a look at a new asynchronous approach. A self-clocked feature makes the design job as simple as the conventional synchronous approach. And you gain a number of advantages:

- Fast response—the circuit does not wait for a clock edge.
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With all these features, the designer has all the leeway of synchronous designs—a minimal number of states and arbitrary input signal changes.

An internal clock is generated when input changes occur. The clock pulse then drives edge-clocked flip-flops that change the state. In this sense internal operation is similar to that of synchronous systems. This overcomes the traditional asynchronous bugaboos of races, hazards and the need to assign extra external states to combat them. In addition there needn't be any restrictions against multiple-input changes.

The conventional asynchronous machine is usually represented by combinatorial logic plus a delay section (Fig. 1). The delay block represents internal propagation delays through the combinatorial logic. Any change of input variables results in an immediate change in state variables, outputs or both.

With the restriction on multiple input changes, Table 1 could represent a typical asynchronous machine with four states, A, B, C and D. The circled letters represent stable states; the others transitions to new stable states. To avoid race conditions, a common requirement for asynchronous design is that only one input variable change at any instant of time. Don't-care assignments are used for multiple changes.

A critical race exists for a transition between stable states A to C for input $x_1 x_2 = 10$. State

A (00) must change to C (11). In so doing state B (01) might occur and the ultimate state attained is stable state D (10). Extra states can avoid races through the adjacent state transitions they provide.¹ Or use edge-sensitive flip-flops and avoid the need for additional states.

One scheme (Fig. 2) allows arbitrary state assignments within the restriction of single input transitions.² An input change produces a $0 \rightarrow 1$ or $1 \rightarrow 0$ transition at the output of the block labeled $f(x)$. The delay element—triggered by $f(x)$ on either transition—clocks the flip-flops at a later time, T. Practical circuit elements include a modulo-2 adder (whose output equals ONE for an odd number of inputs) followed by a one-shot for the delay element.

Delay T is made long enough to ensure proper setup time for the flip-flops and to allow for propagation delays through the system. Thus the delay block of Fig. 1 now becomes a controlled delay (as in the synchronous case), but the combinatorial logic in Figs. 1 and 2 remains the same.

At first glance, the flip-flop method appears costly. But you can use the minimum number of state variables with comparatively few flip-flops. In the true asynchronous example, an extra state variable is necessary. With many input variables, a race problem might not show up until the entire circuit is constructed.

Use a more general design

An extension of the clocked flip-flop approach allows for multiple changes.³ Edge-sensitive flip-flops acquire the state variables, so their number remains minimal. Rather than generate a clock for arbitrary input change, we now produce a trigger only when a change of input warrants a change of internal state. The clock is made a function of the input variables, together with the state variables, to furnish the selective trigger.

The combinatorial logic provides the state variables, as in Fig. 2, but does not provide the output logic (Fig. 3). The state flip-flops perform their original functions; however, the clock input (derived from Block 3) depends on both

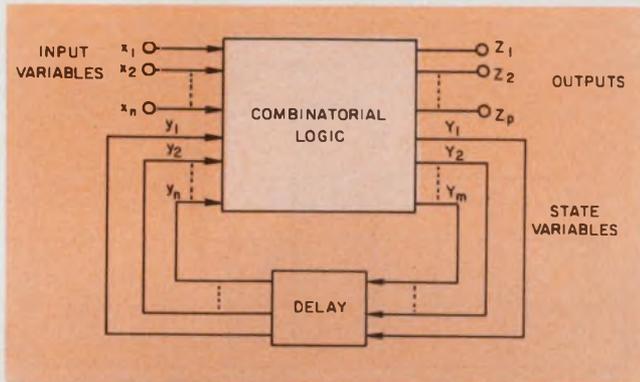
Michael J. Charland, Development Engineer, Canberra Industries, Inc., 45 Gracey Ave., Meriden, Conn. 06450.

state and input variables. The input to the delay-block must be present and stable for a specified time before an output is issued to the flip-flops.

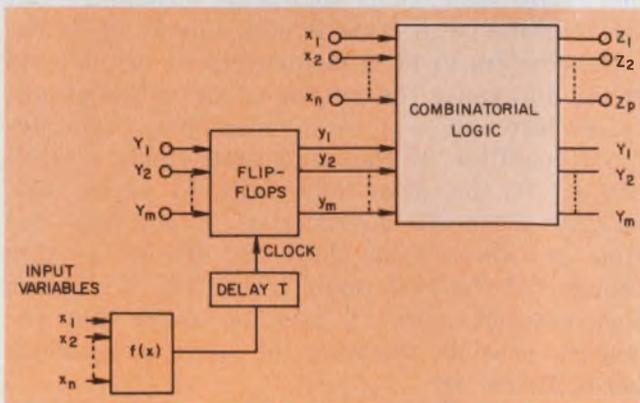
Block 4, along with associated flip-flops, prevents output transients. State variables y_j do not change instantaneously, as inputs x_i can. To avoid possible transients, the output flip-flops receive their clock signal at time T_0 after the clock pulse to the state-variable flip-flops. And the outputs are now hazard-free.

Table 1. Flow table with no multiple input change

| STATE / $y_1 y_2$ | INPUTS $x_1 x_2$ | | | |
|-------------------|------------------|-----|-----|-----|
| | 00 | 01 | 11 | 10 |
| A / 00 | (A) | (A) | B | C |
| B / 01 | — | A | (B) | D |
| C / 11 | — | A | B | (C) |
| D / 10 | A | — | B | (D) |



1. Any change of input variables in an asynchronous machine results in an immediate change of state or output variables. The various internal propagation delays are represented by a single delay block.



2. Use of edge-triggered flip-flops permits arbitrary state assignments with asynchronous design. Any change of input variable sets off the clock, but also allows time for the original state variables to settle.

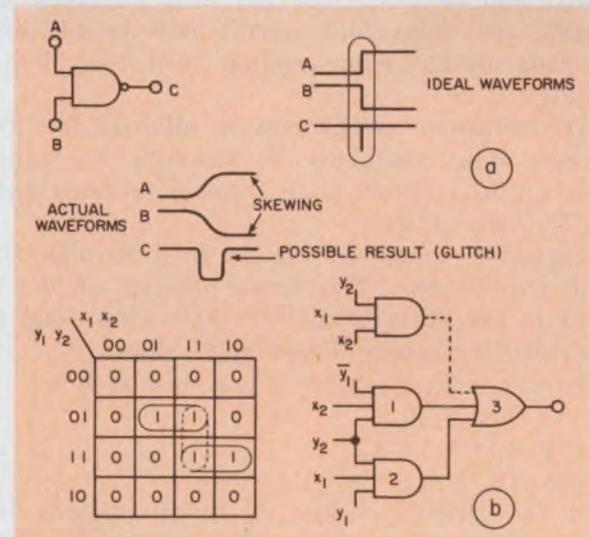
A majority element plus a delay circuit produces the delay-hold effect. To react, the circuit requires the presence of an input for a time, T (Fig. 4).

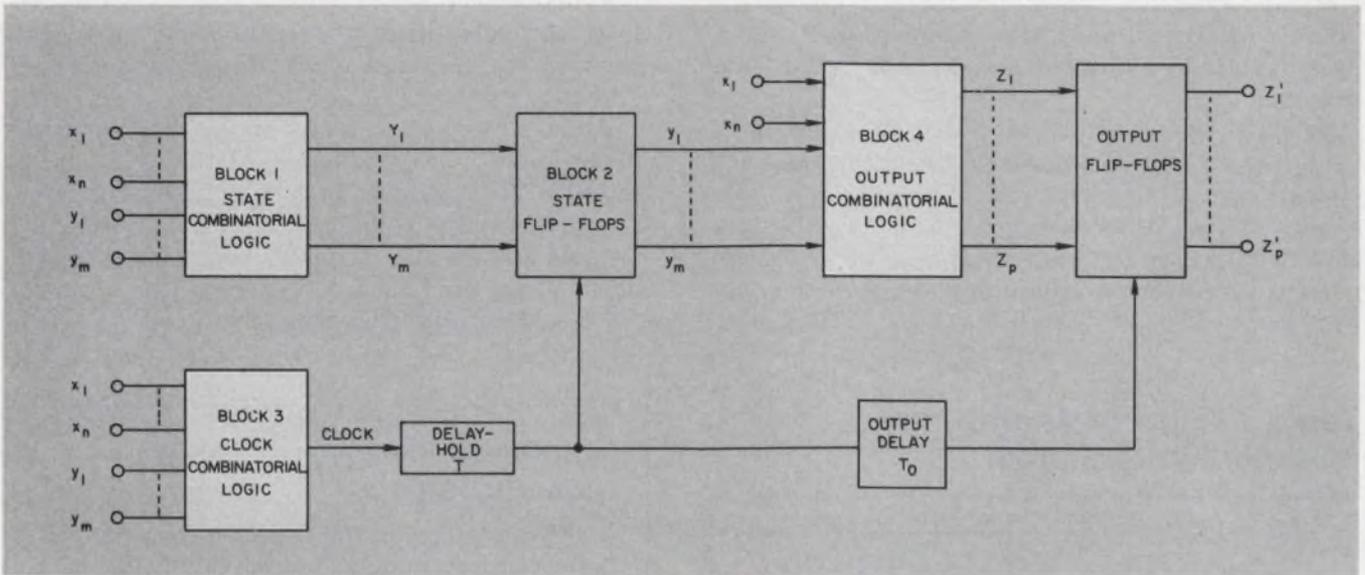
Table 2 describes a sequential machine with double input changes. A map of the clock circuit provides for a pulse only for transition to a new stable state. To obtain this map, insert 0 for a stable state and a 1 for an unstable state. The ONE output produce a $0 \rightarrow 1$ transition from the delay-hold circuit. And when the machine reach-

Conventional asynchronous machine pitfalls

Skewing—the simultaneous change of gate inputs—can lead to an erroneous output transient or glitch (a) because the asynchronous design consists mainly of logic gates whose outputs are fed back to the input. Any erroneous output glitch could be recognized at the input and cause the machine to acquire a wrong state or even step through a number of states. This is a main source of problems in an asynchronous design and, as can be imagined, very difficult to troubleshoot.

Outputs that do not return to the input need no restriction, since they do not control internal sequencing. If, however, the transients due to skewing are harmful to the output they control, then they must be eliminated. Output transients or glitches are called hazards. They can be eliminated by a double cover of the minterms in the output function Karnaugh Map. This type of additional cover is required when adjacent minterms are not completely covered (b). The dotted encirclement adds the top gate and ensures that the output remains high as it should (free of hazards) while the lower gates switch. It is basically for the problems of critical races and hazards that the triggered flip-flop methods have evolved.





3. Selective timed clocks, triggered by input and output variables, remove single-input change restrictions. Use of

output flip-flops, in conjunction with a controllable delay, prevents logic hazards.

Table 2. Flow table with double input change and decomposition to clock and state tables

| STATE / $y_1 y_2$ | INPUTS $x_1 x_2$ | | | |
|-------------------|------------------|-----|-----|-----|
| | 00 | 01 | 11 | 10 |
| A / 00 | (A) | B | D | (A) |
| B / 01 | (B) | (B) | (B) | D |
| C / 11 | B | (C) | B | A |
| D / 10 | A | C | (D) | (D) |

FLOW TABLE

| STATE | INPUT | | | |
|-------|--------------|----|----|----|
| | $x_1 x_2$ 00 | 01 | 11 | 10 |
| A | 0 | 1 | 1 | 0 |
| B | 0 | 0 | 0 | 1 |
| C | 1 | 0 | 1 | 1 |
| D | 1 | 1 | 0 | 0 |

CLOCK CIRCUIT TABLE

| STATE | INPUT | | | |
|-------|--------------|----|----|----|
| | $x_1 x_2$ 00 | 01 | 11 | 10 |
| A | — | B | D | — |
| B | — | — | — | D |
| C | B | — | B | A |
| D | A | C | — | — |

STATE TRANSITION TABLE

es a new stable state, the clock will return to ZERO and be ready for the next stimulus. Of course, the delay-hold circuit will ignore any spurious inputs to its terminal until delay T has ended.

An arbitrary assignment is allowed for unclocked state variables Y_1 through Y_m , since clock outputs result from input conditions that call for new states.

Karnaugh maps (Table 3) help furnish the logic expressions. The large number of don't-cares in the state transition table often lead to inexpensive circuits. These maps show

$$\text{Clock} = y_1 y_2 x_2 + y_2 x_1 x_2 + y_1 x_1 x_2 + y_1 y_2 x_1 + y_1 y_2 x_1,$$

$$Y_1 = \bar{y}_1 x_1 + y_1 \bar{x}_1 x_2,$$

$$Y_2 = y_2 (x_1 + x_2) + x_1 x_2.$$

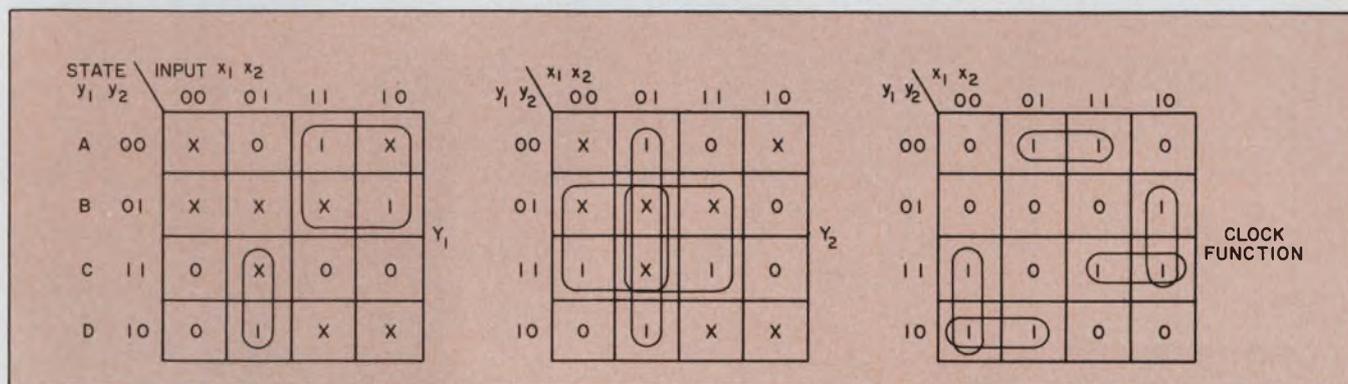
In the circuit of Fig. 4, input changes influence the state variables after a set delay, following which a clock is applied to the flip-flops.

If the hold delay becomes too short, critical races and hazards loom, particularly with multiple input changes. Consider the possible functional hazard for insufficient hold delay (Table 2). Let the machine be in state B with inputs 00. If the input changes to 11, a momentary 01 or 10 could result and cause the output of the clock circuit to produce a $0 \rightarrow 1 \rightarrow 0$ pulse. Why? The 10-input condition of the clock map yields a ONE output. If the delay-hold fails to reject this hazard, a circuit malfunction is possible—the state transition table does not specify a state change for the real input transition. To ensure sufficient delay, time T must be larger than the longest possible duration of such a spurious pulse. Therefore

$$T \geq T_1 = t_{1(\max)} - t_{3(\min)} + d, \quad (1)$$

where $t_{1(\max)}$ is the maximum stray delay in Block 1, $t_{3(\min)}$ is the minimum stray delay in Block 3 and d is the maximum time difference

Table 3. Karnaugh maps for states and clock



between input changes.

Possible race problems also exist. The flow table (Table 2) indicates a race condition if the hold delay is too short. Consider stable state A with input 10. Allow the input to change to 11, and the clock generates a 0 → 1 transition—which gets through the delay-hold—and then triggers the state variable flip-flops. Now, instead of a clean transition to state D, there is a possible pass through state B because of races. This could cause multiple 0 → 1 → 0 transitions, as seen in the clock map. To avoid this type of malfunction, another requirement for T can be established:

$$T \geq T_2 = (\omega_{\max} + t_{3(\max)}) - (\omega_{\min} + t_{3(\min)}), \quad (2)$$

where ω_{\max} = the maximum switching time of the flip-flop,

ω_{\min} = the minimum switching time of the flip-flop,

$t_{3(\max)}$ = the maximum stray delay in Block 3,

$t_{3(\min)}$ = the maximum stray delay in Block 3.

A further requirement for T is it must allow for the proper setup time of the flip-flop. This must include a deduction for the minimum propagation time of the clock generation.

With these constraints,

$$T \geq T_3 = (t_{1(\max)} + d + \epsilon) - t_{3(\min)}, \quad (3)$$

where $t_{1(\max)}$, d, $t_{3(\min)}$ have the same meaning as before and ϵ is the maximum flip-flop setup time.

It is easily seen from Eqs. 1, 2 and 3 that if

$$T \geq \text{MAX}(T_1, T_2, T_3), \quad (4)$$

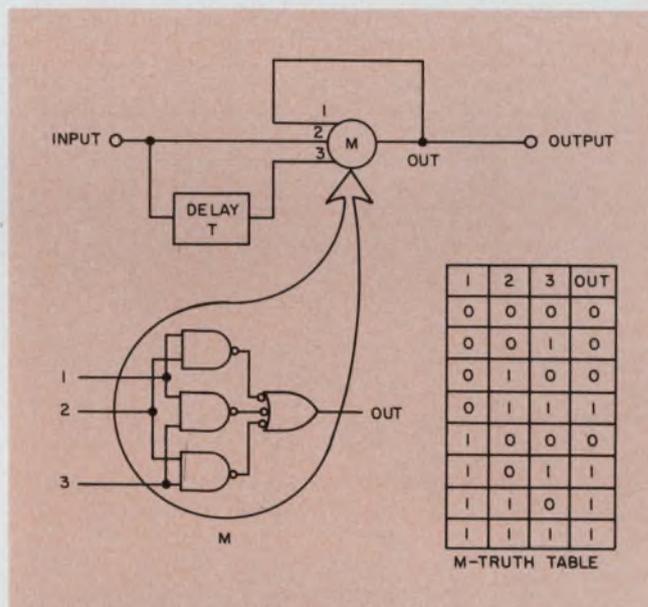
then the delay-hold not only smooths the transients caused by hazards, but also guarantees that when the trigger pulse arrives at the flip-flop clock inputs, all the data lines are stable.

T_3 is obviously greater than T_1 , so that Eq. 4 is equivalent to

$$T \geq \text{MAX}(T_2, T_3). \quad (5)$$

With most TTL J-K flip-flops, T_2 is definitely smaller than T_3 and Eq. 5 further reduces to

$$T \geq T_3 = (t_{1(\max)} + d + \epsilon) - t_{3(\min)}. \quad (6)$$



4. Majority logic plus delay provides the delay-hold circuit. The purpose: ignore inputs with changes that occur during a predetermined interval T.

Delay T_0 in Fig. 3 is needed to ensure enough time for the state variables to propagate through Block 4 and meet the setup time of the output flip-flops. Thus

$$T_0 \geq \omega_{\max} + \epsilon + t_{4(\max)},$$

where ω_{\max} = the maximum switching time of the state-variable flip-flops.

ϵ = the maximum flip-flop setup time,

$t_{4(\max)}$ = the maximum stray delay in Block 4.

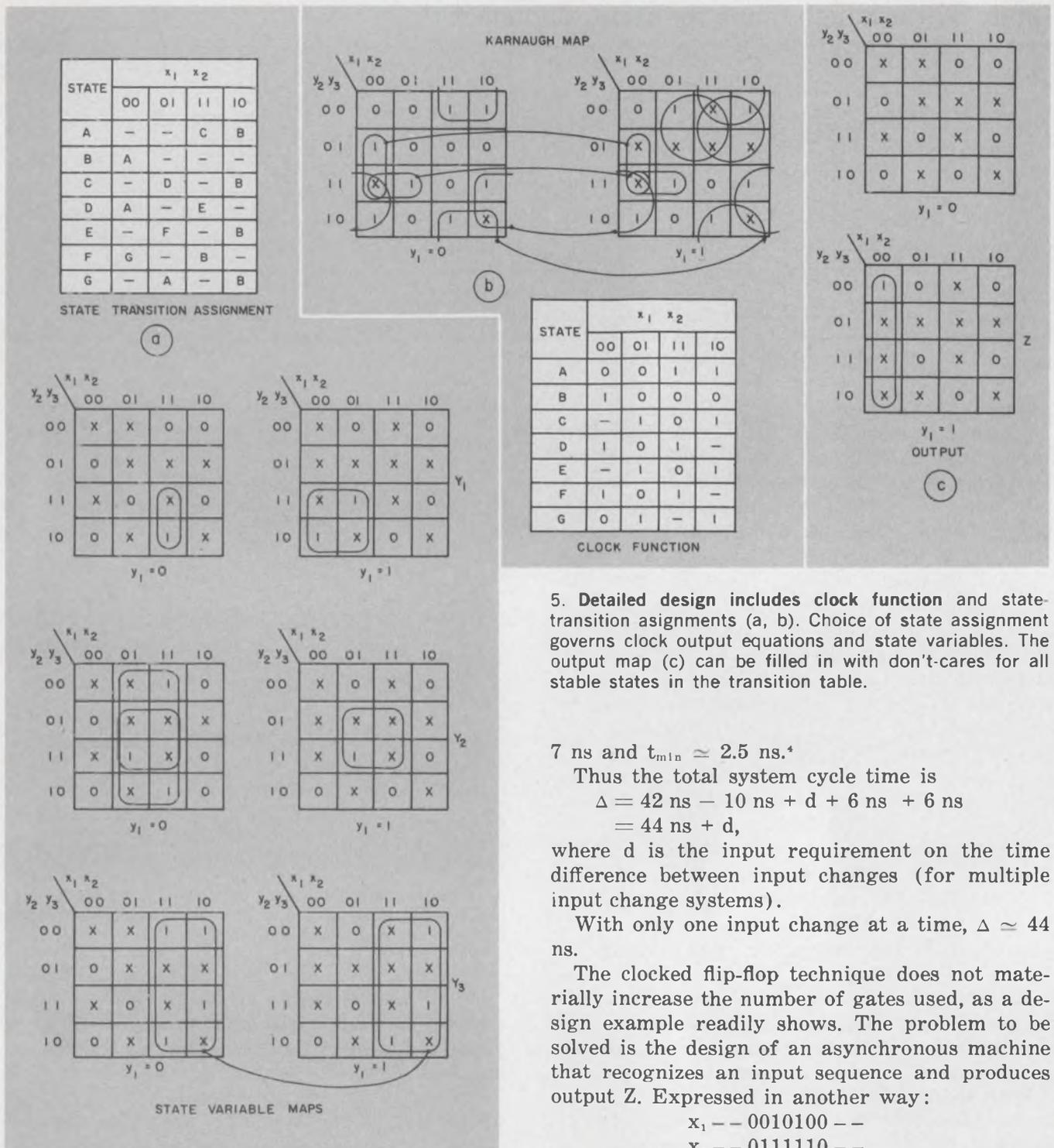
The time for one cycle of operation is the time between any two input stimuli that will cause the circuit to react. If we disregard the output, the cycle time for state change is

$$\Delta' = t_{3(\max)} + T + \omega_{\max}. \quad (7)$$

Including the output function, the total cycle time is

$$\Delta = t_{3(\max)} + T + T_0 - t_{4(\min)}. \quad (8)$$

The addition of $T_0 - T_{4(\min)}$ to the state cycle time is needed to ensure that the x inputs are



5. Detailed design includes clock function and state-transition assignments (a, b). Choice of state assignment governs clock output equations and state variables. The output map (c) can be filled in with don't-cares for all stable states in the transition table.

7 ns and $t_{min} \approx 2.5$ ns.⁴

Thus the total system cycle time is

$$\Delta = 42 \text{ ns} - 10 \text{ ns} + d + 6 \text{ ns} + 6 \text{ ns} = 44 \text{ ns} + d,$$

where d is the input requirement on the time difference between input changes (for multiple input change systems).

With only one input change at a time, $\Delta \approx 44$ ns.

The clocked flip-flop technique does not materially increase the number of gates used, as a design example readily shows. The problem to be solved is the design of an asynchronous machine that recognizes an input sequence and produces output Z. Expressed in another way:

$$\begin{aligned} x_1 & \text{ -- 0010100 --} \\ x_2 & \text{ -- 0111110 --} \\ Z & \text{ -- 0000001 --} \end{aligned}$$

Three state variables suffice, but four ensure race-free operation with the conventional asynchronous design outlined in Fig. 1. The equations for states and outputs are:

State variables:

$$\begin{aligned} Y_1 & = \overline{x_1} + \overline{x_2} + \overline{y_3} + \overline{y_4} \\ Y_2 & = \overline{x_1} + y_3 x_2 + \overline{y_3} \overline{y_4} + y_3 \overline{y_4} \\ Y_3 & = \overline{x_2} + x_1 y_3 + x_1 \overline{y_2} + y_1 y_3 \\ Y_4 & = x_1 x_2 + x_1 y_3 + x_2 y_3 y_4 \end{aligned}$$

Output:

$$Z = \overline{y_2} y_3 \overline{x_1} \overline{x_2}$$

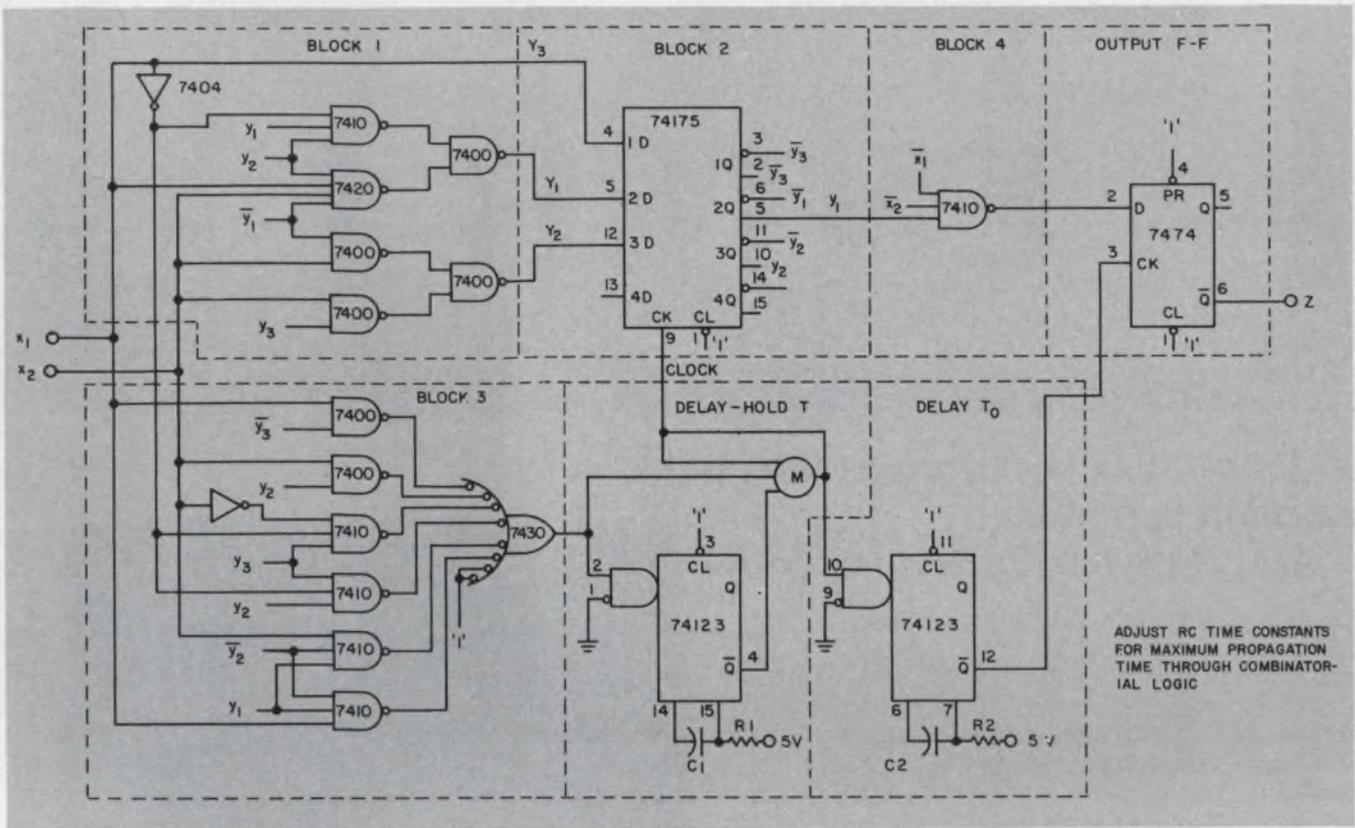
The triggered flip-flop method will start with

stable and do not cause an output change prior to clocking the output flip-flops.

For instance, with two-level combinatorial logic for all blocks and $t_{i(max)} = 2 t_{(max)}$; $t_{i(min)} = 2 t_{(min)}$ ($i = 1, 3, 4$) and where t_{max} and t_{min} are the corresponding gate delays, then

$$\begin{aligned} \Delta & = 2 t_{max} + (2 t_{max} + d + \epsilon - 2 t_{min}) \\ & \quad + (\omega_{max} + \epsilon + 2 t_{max}) - 2 t_{min} \\ & = 6 t_{max} - 4 t_{min} + d + 2 \epsilon + \omega_{max} \end{aligned}$$

For a Schottky J-K flip-flop, $\epsilon \approx 3$ ns and $\omega_{max} \approx 6$ ns; so for typical gates of the same line, $t_{max} \approx$



6. Completed circuit shows clearly defined segments related to the block diagram of Fig. 4. Two one-shots

add the necessary delays and permit individual adjustment for hold and output delay times.

the conventional reduced flow table, rewritten as follows:

| State | $y_1 y_2 y_3$ | $x_1 x_2$ | | | | Z |
|-------|---------------|-----------|-----|-----|-----|---|
| | | 00 | 01 | 11 | 10 | |
| A | 000 | (A) | (A) | C | B | 0 |
| B | 001 | A | (B) | (B) | (B) | 0 |
| C | 011 | - | D | (C) | B | 0 |
| D | 010 | A | (D) | E | - | 0 |
| E | 111 | - | F | (E) | B | 0 |
| F | 110 | G | (F) | B | - | 0 |
| G | 100 | (G) | A | - | B | 1 |

Proceed, as before, to draw up maps of the clock and state-transition tables. The clock output need only be ONE for a transition to a new state. Therefore all stable states in the flow table are don't-cares, since there will be no clock output to effect a change. The arbitrary state assignment chosen (Fig. 5a) determines the clock output equations (Fig. 5b) and the state variables. The single output is derived in Fig. 5c. The

output map can also be filled in with don't-cares for all stable states, since there is no clock output for these states.

The following expressions are taken from the maps:

$$Y_1 = y_1 y_2 \bar{x}_1 + \bar{y}_1 y_2 x_1 x_2$$

$$Y_2 = y_3 x_2 + \bar{y}_1 x_2$$

$$Y_3 = x_1 \bar{x}_2$$

$$Z = y_1 x_1 \bar{x}_2 \bar{y}_2$$

$$C = y_2 x_2 + \bar{y}_3 x_1 + y_1 \bar{y}_2 x_1 + y_1 \bar{y}_2 x_2 + y_3 \bar{x}_1 y_2$$

Compared with normal asynchronous design, there is no need to work with an adjacency diagram and introduce another state variable. A more careful choice of state assignment can realize the clock function with still fewer gates. To complete the design add the delay-hold circuit, the T_0 delay, three state variable flip-flops and one output flip-flop (Fig. 6). ■■

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1. Smith, W.R., "Level Operated Sequential Machines," Course CE 485 handout, University of Bridgeport, Bridgeport, Conn. Spring Semester 1974, pp. 8-6 and 8-7.
2. Bredeson, J.G., and Hulina, P.T., "Generation of a Clock Pulse for Asynchronous Sequential Machines to Eliminate Critical Races," IEEE Trans. Comput., Vol. C-20, February, 1971, pp. 225-226.
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ENTER ↑

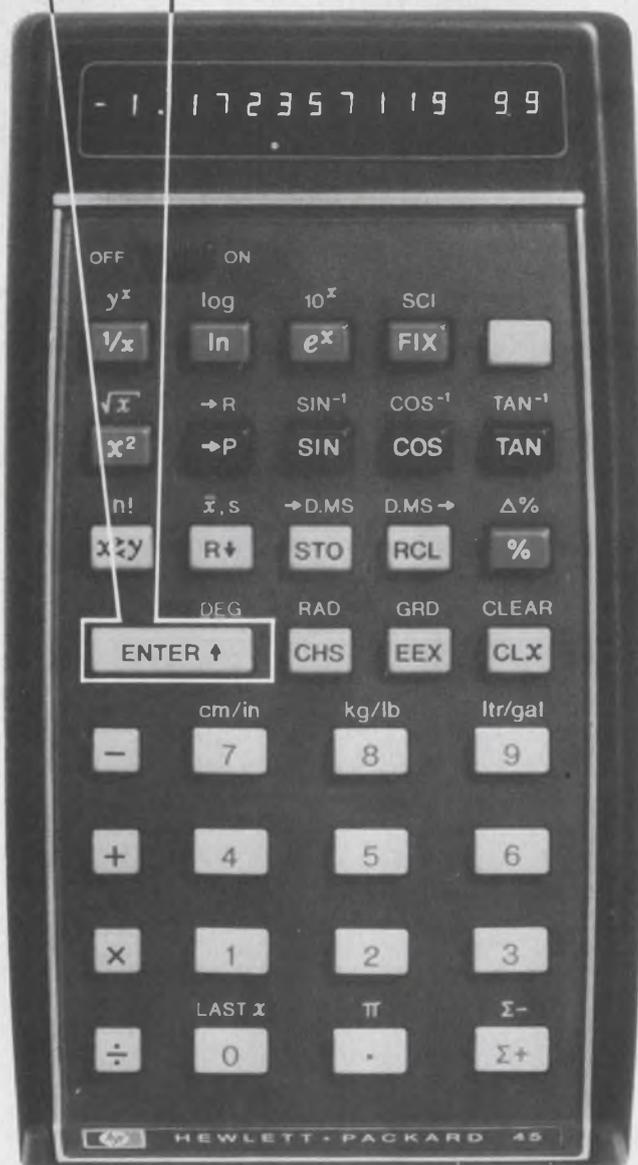
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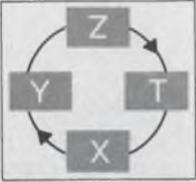
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4. You almost never have to re-enter intermediate answers—a real time-saver, especially when your data have eight or nine digits each.
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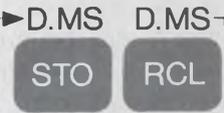
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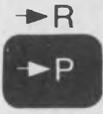
2.  It offers a 4-Register Operational Stack that saves intermediate answers and automatically retrieves them when they are required in the calculation.

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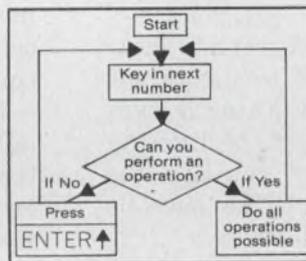


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Is a crowbar alone enough? To reduce the risk of power-supply damage or even fires, an overcurrent interrupt element may also be needed.

Even with a crowbar, a low-voltage, series-regulated power supply may still destroy itself, harm your load or—under extreme circumstances—touch off a fire.

This can happen because, in the course of its sole job—to protect a load from overvoltage—the crowbar draws short-circuit currents. These currents, if not limited, can lead to dangerous secondary heating. While current limiting is built into most crowbar supplies, what if the limiting action doesn't work? Don't worry, you say, the fuse is sure to blow—eventually. This may be true. But if "eventually" isn't fast enough . . .

To counteract both possibilities, an overcurrent interrupt element should be tied in to the crowbar. Then if the crowbar trips—and the limiting or the fuse doesn't function—the overcurrent interrupt element will shut down the supply.

A crowbar alone isn't enough

With both the crowbar and interrupt element in the supply, you can rest easy. After all, your load is now fully protected.

Or is it?

Just as the presence of a crowbar doesn't by itself ensure against all hazards, a crowbar's mere presence doesn't necessarily guarantee adequate overvoltage protection: All crowbars are not alike.

Therefore look for the following characteristics in a well-designed crowbar circuit:

- Comparator-circuit bias—should be independent of the main power and reference circuitry.
- Response speed—can't be too slow—or too fast.
- Fuse selection—must ensure a fuse blow at the right time.
- Crowbar thyristor location—must be selected for optimum protection.

If current limiting doesn't function, and the line fuse doesn't open when the crowbar fires, you can expect rapid overheating of the crowbar thyristor, of any isolation diodes in series with the thyristor, and of a number of resistors within the supply. The latter includes resistors for current monitoring, series-regulator emitter equalizing and for surge limiting.

How much overheating can you expect? If the initial fault, say, is a series-regulator short, then the emitter resistor will be forced to dissipate at least twice its normal amount. This high dissipation may melt the coating of wirewound resistors or—carried to the extreme—reach the flash point of nearby materials.

Overheating in the crowbar thyristor or any of its isolating diodes may result in a strange phenomenon. The components neatly unsolder themselves from the circuit. If this happens, the crowbar can no longer clamp the output, and the voltage at the load rises to that of the supply's unregulated dc level. What happens to the load? Poof!

To ensure a fuse blow, some supplies connect the crowbar thyristor across the input-rectifier filter capacitor and the crowbar sensing circuit across the output terminals (Fig. 1). When the thyristor fires, it discharges the capacitor and clamps the rectifier voltage to a low value. If the current path through the thyristor is of low-enough impedance, the overload will reflect back to the ac line and open the fuse.

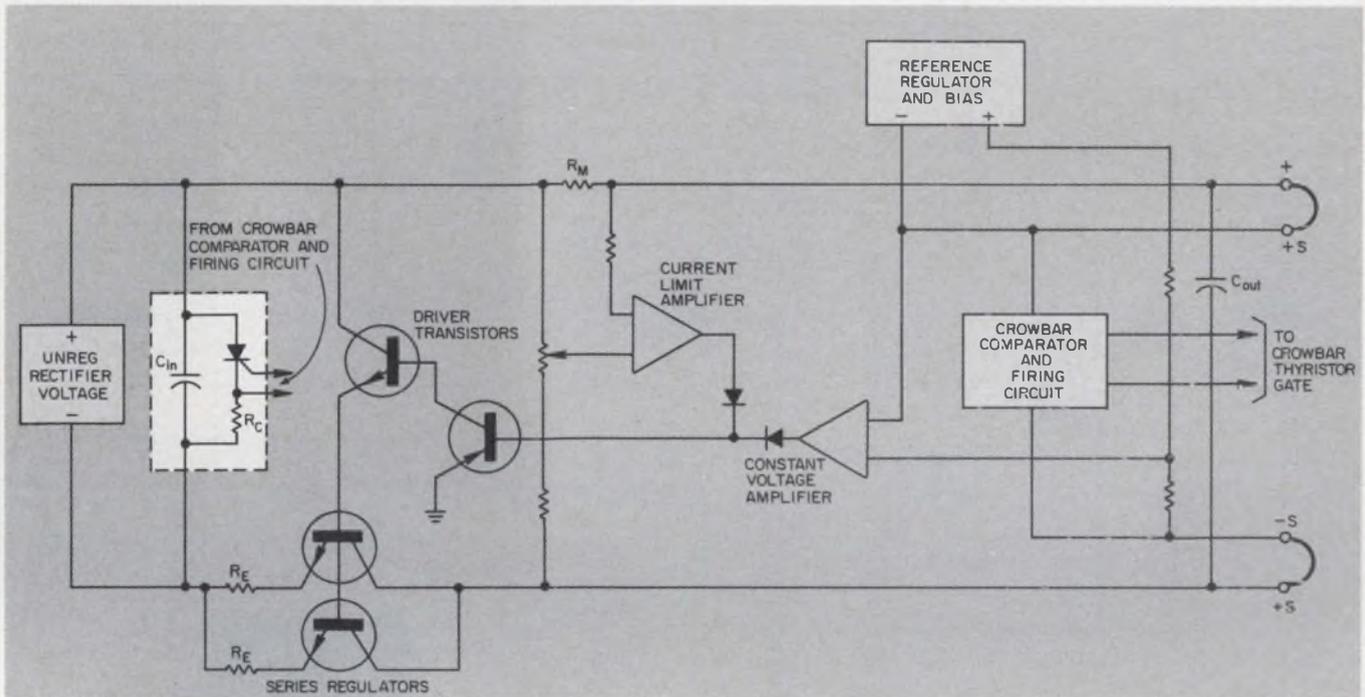
Where's the fuse?

There are several drawbacks to this configuration, however. First, the ac line fuse will probably—and annoyingly—blow more frequently. This means that an abundant supply of fuses must be readily available to avoid needless downtime.

If the power supply is also buried in an all but inaccessible location—forget it. (Don't forget, however, that frequent manual fuse changes pre-empt the advantages of automatic-reset crowbars.)

Another limitation of the "inboard" crowbar

Willis C. Pierce, Jr., Project Manager, Hewlett-Packard, New Jersey Division, Green Pond Rd., Rockaway, N.J. 07866.



1. To ensure a fuse blow when a fault occurs, some power supplies place the crowbar thyristor across the

input-rectifier filter capacitor. But this may cause the fuse to blow more frequently.

What is a crowbar?

In a crowbar, when a power supply's output voltage rises above a predetermined, acceptable level, circuitry detects the overvoltage condition and fires a thyristor (SCR or triac) connected across the power-supply output terminals. The thyristor places a virtual short across the terminals and thereby clamps the output voltage to a safe level.

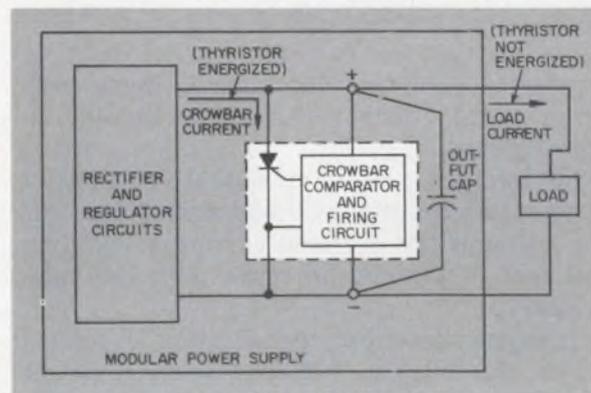
Whether SCR or triac, the significant characteristic of this element is that it must behave as a switch. When the crowbar is in its non-actuated state, the element is a virtual open circuit that draws no current from the power supply. When the crowbar is actuated, the thyristor switches on and begins to absorb the output current while it forces the output voltage down (see figure).

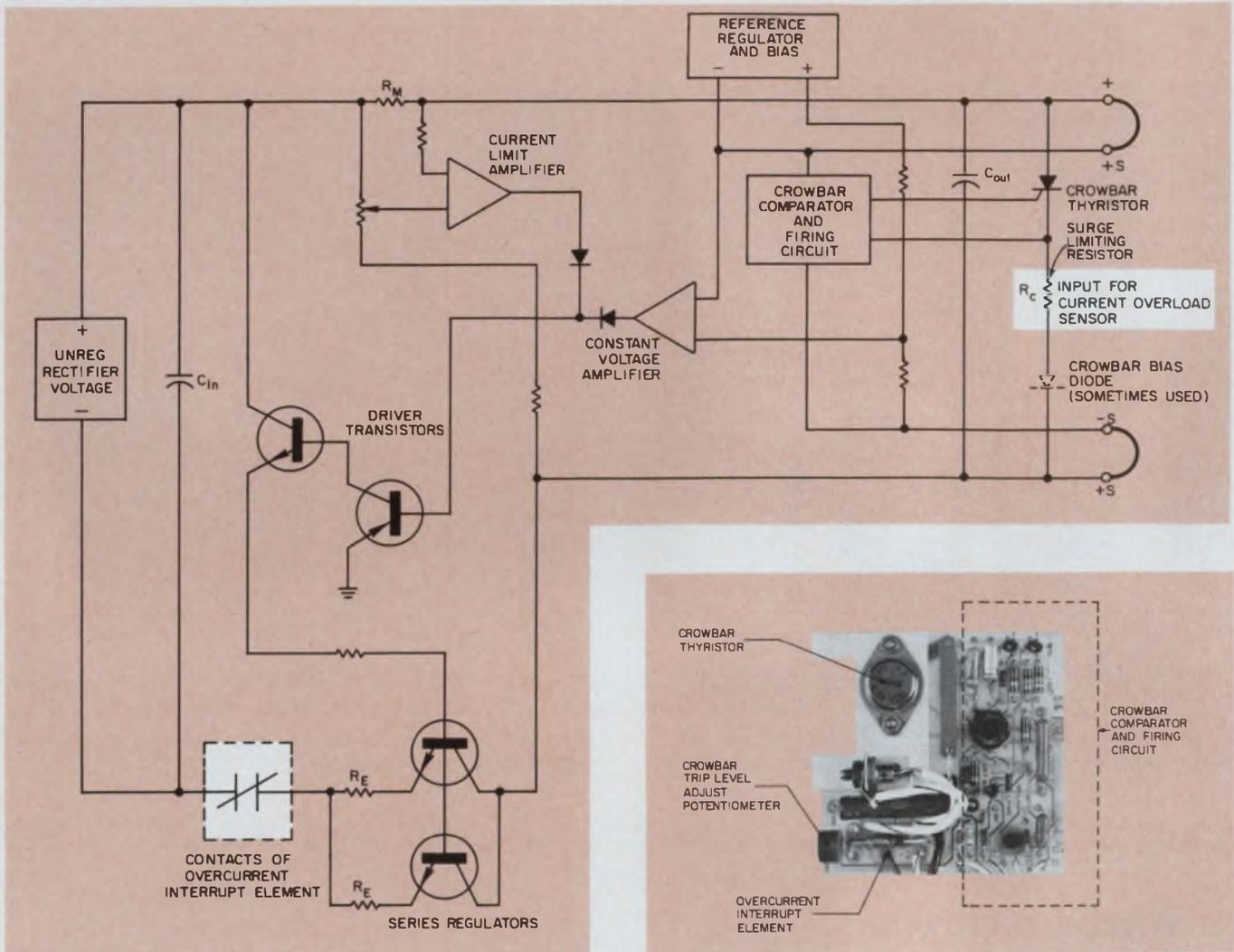
The crowbar thyristor is controlled by a circuit that compares the power-supply output voltage to that of a reference. As long as the output is less than the reference, no firing signal is sent to the thyristor. When the output exceeds the reference, the comparator signals the thyristor, which then switches on.

The supply output voltage is held to the thyristor's saturated forward drop—1 to 1.5 V.

When the supply senses the increase in load current, it goes into the foldback current mode to limit the internal dissipation and that of the thyristor.

The power supply and crowbar remain in a state of equilibrium until input power is removed. The crowbar then resets, and the supply returns to normal operation after input power is reapplied. The sequence of operation from firing to current limit requires only tens of milliseconds before steady state is reached.





2. To protect against damage if both the fuse and the foldback-current limiting circuit become inoperative, an overcurrent interrupt element is used along with the

crowbar. When a fault occurs, the interrupt element's contacts open the main series power path. With this method, both the supply and load are fully protected.

is the stress placed on the input rectifiers when the crowbar fires. Since the crowbar thyristor places a virtual short across the unregulated dc, the rectifiers must provide the overload current until the ac line fuse opens and interrupts the input power. The repeated overload surges can adversely affect rectifier life.

ly long time constant to prevent false actuation of the protection device. This can happen when the crowbar initially fires and receives a surge from the output-capacitor discharge.

In supplies that use foldback current limiting, the preferred solution—one that protects both supply and load—rests with the overcurrent interrupt element (Fig. 2).

Thus the protection device must be selected so the thermal time constant of the heater element is significantly shorter than that of the other circuit elements. But the time constant must be long enough to prevent tripping of the overload device when the crowbar operates upon an external failure, rather than an internal one.

In this method, a heater resistance element in a nonresettable thermal-overload device uses the heating value of the overload current to open an isolated pair of contacts in series with the main power path.

The non-resettable, or one-shot, characteristic of the interrupt element prevents the power supply from being turned on after a fault has occurred. This stops repeated on-off cycling of the supply, which can result in unnecessary stress on internal components, or of the load.

The requirements for the sensing/protection device are as follows: (1) The sensing circuit must be electrically isolated from the contacts; (2) The interrupt element must have a sufficient-

To replace the actuated overload device presents no particular problem since, presumably, the device opened because of an internal failure

and the power supply must be disassembled anyway for repair.

Since the thermal overload actually operates at some fixed value of temperature, and since the device is heated by the current through its resistor, both the ambient temperature and ac line voltage will influence the trip point.

Select for worst case

Therefore the interrupt element must be selected or biased with additional resistors to ensure satisfactory operation under worst-case conditions of ambient and line voltage. One worst-case condition occurs at low line voltage and low ambient temperature. This condition results in the longest actuating time.

But while the device must actuate fast enough to protect at low line, low ambient, the actuating time must be long enough to prevent false tripping. This can occur under high ambient conditions or when the crowbar actuates but no power-supply failure has occurred.

Though the resistively heated thermal-overload is probably the best way to provide secondary protection, other methods exist. For example, reed relays or Hall-effect sensors can be used to sense the overload current magnetically. But these require auxiliary circuitry at additional cost.

Another method uses an electromechanical thermal device, such as a thermostat. The thermostat, which can be automatically or manually reset, operates as the one-shot thermal overload does, but doesn't respond as fast. Also it is larger and more expensive.

Note that, with normal crowbar action, the element will not actuate. When the crowbar trips—and there has been no power-supply failure—the thyristor clamps the output voltage to a low value. The power-supply current limit then takes over and the supply remains in equilibrium until the ac input power is removed and reapplied.

If, however, the crowbar trips because of an internal failure, and the current limit circuit is not operating, then the interrupt element shuts down the supply before a serious hazard or load damage occurs. Thus the overload-protected crowbar may be considered as one with secondary protection.

To avoid the possibility that a single component failure can disable both the power supply and its crowbar, the crowbar's comparator circuit should be biased separately from the rest of the power supply. The crowbar may derive power from the supply transformer windings and rectifiers, but should have its own reference element, comparator and firing circuitry.

Crowbar circuits are sometimes built as two-terminal devices to be connected across the power-supply output terminals. In these circuits, the crowbar reference and comparator circuits undesirably draw bias and excitation power from the power-supply output.

Also undesirable: And the load may not be adequately protected if remote sensing is used. This is because the crowbar's trip level must set high enough to compensate for both normal drops in the load leads and output transients caused by unloading. The sum of these effects can be enough to permit excessive load voltage before the crowbar senses the fault.

By contrast, a four-terminal crowbar permits the crowbar thyristor leads to be connected at the output terminals, while the crowbar sense leads can be connected to the remote sense terminals. Thus the crowbar senses the actual load voltage and is not affected by the drop across the load leads.

Critical crowbar specs

Another critical parameter is the crowbar's speed of response to an overload. If the crowbar's response time—the time required to sense and trigger following an overvoltage—is too fast, then spurious noise can cause false tripping. If the response is too slow, then the output can rise high enough to damage the load.

As an example, assume that a 5-V, 16-A power supply fails because of a short in one of its series-regulator transistors. Now with 10-V across a 40,000- μ F rectifier filter capacitor, an output capacitor of 10,000 μ F, a load voltage of 5 V and 17 m Ω of resistance in series between the two capacitors, the voltage across the load rises to 5.5 V in 17.9 μ s.

Because of the crowbar's trip margin, which may be as much as 1 V above the output-voltage setting, and because of the rapid rate of rise of the output voltage, the crowbar should respond in about 20 μ s. This will ensure safe limits for TTL and similar loads.

Safety in power supplies frequently depends on the fuse—that thin string of metal whose only function is to burn up and die. In one of the most common power-supply failure modes—that of a shorted series regulator—current limiting becomes inoperative and the burden of protection falls upon the fuse. Unfortunately the fuse may not want to die. Here's why.

Don't depend on the fuse

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To do this, the power-supply engineer designs the power transformer and filter capacitor to provide just enough dc voltage to prevent saturation of the series-regulator transistors under worst-case operating conditions. This means that the low-line, full-load rectified input voltage may be only a few volts higher than the regulated output voltage.

When the thyristor fires and clamps the output voltage, it produces a saturated drop of between 1 and 1.5 V across the power-supply output terminals. This voltage subtracts from the available unregulated rectifier voltage. The dc current is then determined by the rectified voltage less the thyristor voltage, divided by the sum of the series circuit resistances. Included in series are the resistors for current monitoring and emitter equalizing, the thyristor surge-limiting resistor, and the resistance of the circuit-board conductors and interconnecting wires.

Though the sum of these series resistances seems to be rather low, the total is not much less than that which the regulator's dc load resistance presents to the input rectifier filter capacitor under normal operation. Thus thyristor current under fault conditions is limited by the power-supply circuitry, and this current may be sufficient to blow the fuse.

Fault current is limited by another factor—the regulation characteristic of the unregulated dc supply (transformer, rectifier, filter capacitor) under fault conditions. When a fault occurs in the dc load circuitry, the output voltage rises, the crowbar fires and drastic changes occur in the operation of the rectifier filter capacitor. In effect, the rectifier input sees a reduction in load resistance, which results in a lower average voltage at the capacitor terminals. Consequently the overload current is limited to a value that can be insufficient to open the ac line fuse.

Typically, the rectifier capacitor sees an apparent load of 1.08 Ω when the power supply behaves normally, and 0.36 Ω when one of the series-regulator transistors shorts. The rms line current changes from 1.4 to only 2.4 A under fault conditions even though the load impedance changes by a factor of three to one. If the power supply carries a 2.5-A fuse—probably the minimum value necessary to avoid annoying fuse blowing upon turn on—then it is obvious that the ac line fuse can't adequately protect the supply.

Fuse selection isn't simple

Therefore to make sure the fuse does its job, it should be selected with these criteria in mind: With a capacitive input filter, the fuse must

handle the steady-state line current, as well as the peak inrush currents that occur on turn on. To do this, it is common practice to select a fuse whose rated current is between 125% and 250% of the supply's rated full-load line current. This usually provides enough overload capability to prevent fuse blowing from aging and fatigue with repeated turn-on overloads.

Next examine the opening time vs overload characteristics of standard commercial fuses such as the commonly used 3AG or 3AB glass type. Known for reliability, low cost and widespread availability, the 3AG or 3AB offer normal, fast or slow-blow characteristics. These characteristics specify the percentage overload vs typical opening time for the first 100 s of overload. Beyond 100 s the characteristic is the same for all three types: Each can sustain a 135% overload for one hour.

Now, recall the example in which the supply's rms line current increased to only 2.4 A under a fault (17% of full rated current). If the ac line fuse for this supply was selected for 175% of rated line current—and it can sustain a 135% specified overload for one hour—the fuse can operate for up to one hour with 236% of full rated current through it. Clearly, this line fuse cannot do its job when a series-regulator transistor or one of the driver transistors shorts.

An obvious solution to this problem is to place a fuse in the dc-load circuit. This appears to be a good solution because a fault gives a greater percentage increase in dc-load current than in the rms primary or secondary load current. But wait!

To move the fuse to the dc side, you've got to find one whose rating just equals the power supply's output-current rating, and that accepts an overload for approximately 100 s before it opens.

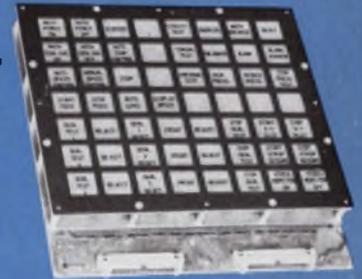
Since fuses normally come in specific values only (they increase in about 1-A increments from 3 to 8 A, 2-A increments from 8 to 12 A and 5-A increments above 15 A), it may not be possible to find one that will open in 100 s under all fault conditions. Even if you do, 100 s can be long enough to permit gross overheating of resistors in the dc circuit and severely damage the supply.

Another important point: Because many possible failure modes exist, and because a failure may not present a zero-impedance short, fault currents in the power-supply load circuit and in the crowbar are not accurately predictable. It's conceivable that a failure mode could limit the fault current and keep it at 135% of the fuse rating, or less. If this were to occur, the fuse would tolerate the overload for much longer than 100 s. The potential problems are obvious. ■■

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SPECIFICATIONS FOR LX SERIES

DC Output

Regulated Voltage

regulation, line 0.1%
regulation load 0.1%
ripple and noise 1.5 mV RMS, 5 mV pk-pk with either
positive or negative terminal grounded

temperature
coefficient 0.03%/°C

AC Input

line 105-132 VAC; 47-440 Hz. For 187-242 VAC, add
suffix "V" to model numbers and add 12% or
\$30.00 to the price, whichever is greater. For
operation of LX-D, LXS-E, LX-EE, LXS-7, and
LXS-8 units at 50 Hz or at 400 Hz, consult factory.
Ratings apply to 57-63 Hz. For all other models
delete 40°C rating for 50 Hz operation.

Ambient operating temperature range

continuous duty from 0° to + 71°C with corresponding load
current ratings for all modes of operation.

Storage temperature range

-55°C to +85°C

Overload protection

Thermal

thermostat, automatic reset when over-temp. condition is re-
moved. (Not applicable to LX-D-3); circuit breaker must be reset
on LX-8 models.

Electrical

external overload protection, automatic electronic current limit-
ing circuit limits the output current to the present value,
thereby providing protection for load as well as power supply.

Overshoot

no overshoot on turn-on, turn-off or power failure.

Input and output connections

through terminal block on chassis; output terminals on LX-7,
LX-8 models are two heavy duty studs.

Power hybrid voltage regulator or integrated circuit regulation

some models have Power Hybrid Voltage Regulator providing
complete regulation system while others have an integrated
circuit providing regulation system except for input and output
capacitors, rectifiers and series regulation transistors.

Controls

DC Output Control

simple screwdriver voltage adjustment over entire range.

Remote sensing

provision is made for remote sensing to eliminate effect of
power output lead resistance on DC regulation.

Transformer

MIL-T-27C, Grade 6

Tracking accuracy (dual models)

2% absolute voltage difference; 0.2% change for all conditions
of line, load and temperature.

Fungus proofing

All fungi nutrient components are rendered fungi inert with
MIL-V-173 varnish—Standard on all LX models and is included
in price.

5 VOLTS ±5% SINGLE OUTPUT

| MODEL | MAX. AMPS AT AMBIENT OF: | | | | PRICE |
|----------------|--------------------------|------|------|------|--------|
| | 40°C | 50°C | 60°C | 71°C | |
| LXS-A-5-OV-R* | 4.0 | 3.4 | 2.7 | 2.0 | \$ 95. |
| LXS-B-5-OV-R* | 5.8 | 5.0 | 4.0 | 3.0 | 130. |
| LXS-4-5-OV-R* | 7.4 | 6.5 | 5.4 | 3.9 | 145. |
| LXS-C-5-OV-R* | 9.0 | 8.0 | 6.8 | 5.3 | 160. |
| LXS-CC-5-OV-R* | 16.0 | 14.5 | 12.7 | 10.5 | 220. |
| LXS-D-5-OV-R* | 27.5 | 24.2 | 20.5 | 16.5 | 260. |
| LXS-E-5-OV-R* | 35.0 | 30.0 | 24.0 | 17.5 | 320. |
| LXS-EE-5-OV-R* | 45.0 | 39.0 | 32.0 | 25.0 | 445. |
| LXS-7-5-OV-R** | 65.0 | 56.0 | 46.0 | 35.0 | 535. |
| LXS-8-5-OV-R** | 85.0 | 77.0 | 68.0 | 56.0 | 580. |

*Includes fixed overvoltage protection at 6.8V ±10%

6 VOLTS ±5% SINGLE OUTPUT

| MODEL | MAX. AMPS AT AMBIENT OF: | | | | PRICE |
|----------------|--------------------------|------|------|------|--------|
| | 40°C | 50°C | 60°C | 71°C | |
| LXS-A-6-R | 3.7 | 3.1 | 2.5 | 1.9 | \$ 95. |
| LXS-B-6-R | 5.5 | 4.7 | 3.8 | 2.9 | 130. |
| LXS-4-6-R | 6.6 | 5.8 | 4.8 | 3.5 | 145. |
| LXS-C-6-R | 8.8 | 7.8 | 6.7 | 5.2 | 160. |
| LXS-CC-6-R | 15.2 | 13.8 | 12.1 | 10.0 | 210. |
| LXS-D-6-R | 26.5 | 23.4 | 19.8 | 16.0 | 260. |
| LXS-E-6-R | 34.0 | 29.0 | 23.0 | 16.5 | 320. |
| LXS-EE-6-OV-R† | 42.0 | 36.0 | 30.0 | 22.0 | 445. |
| LXS-7-6-OV-R** | 59.0 | 50.0 | 41.0 | 32.0 | 535. |
| LXS-8-6-OV-R** | 70.0 | 70.0 | 68.0 | 56.0 | 580. |

†Includes fixed overvoltage protection at 7.4V ±10%

12 VOLTS ±5% SINGLE OUTPUT

| MODEL | MAX. AMPS AT AMBIENT OF: | | | | PRICE |
|-----------------|--------------------------|------|------|------|--------|
| | 40°C | 50°C | 60°C | 71°C | |
| LXS-A-12-R | 2.7 | 2.2 | 1.8 | 1.5 | \$ 95. |
| LXS-B-12-R | 3.8 | 3.6 | 3.0 | 2.2 | 130. |
| LXS-4-12-R | 4.4 | 3.8 | 3.1 | 2.5 | 145. |
| LXS-C-12-R | 6.5 | 6.1 | 5.5 | 4.6 | 160. |
| LXS-CC-12-R | 10.5 | 9.4 | 8.2 | 5.0 | 210. |
| LXS-D-12-R | 16.0 | 14.0 | 11.9 | 8.0 | 260. |
| LXS-E-12-R | 21.0 | 18.0 | 15.0 | 12.5 | 320. |
| LXS-EE-12-R | 32.0 | 27.0 | 22.0 | 16.0 | 420. |
| LXS-7-12-OV-R** | 40.0 | 36.0 | 30.0 | 23.0 | 535. |
| LXS-8-12-OV-R** | 50.0 | 45.0 | 40.0 | 34.0 | 580. |

15 VOLTS ±5% SINGLE OUTPUT

| MODEL | MAX. AMPS AT AMBIENT OF: | | | | PRICE |
|-----------------|--------------------------|------|------|------|--------|
| | 40°C | 50°C | 60°C | 71°C | |
| LXS-A-15-R | 2.4 | 2.0 | 1.6 | 1.3 | \$ 95. |
| LXS-B-15-R | 3.2 | 2.8 | 2.5 | 1.5 | 130. |
| LXS-4-15-R | 4.0 | 3.5 | 2.8 | 2.3 | 145. |
| LXS-C-15-R | 6.0 | 5.6 | 5.1 | 4.5 | 160. |
| LXS-CC-15-R | 9.5 | 8.6 | 7.4 | 4.8 | 210. |
| LXS-D-15-R | 14.0 | 12.3 | 10.4 | 7.5 | 260. |
| LXS-E-15-R | 19.0 | 17.0 | 14.0 | 12.0 | 320. |
| LXS-EE-15-R | 28.0 | 24.0 | 19.5 | 14.0 | 420. |
| LXS-7-15-OV-R** | 36.0 | 32.0 | 26.0 | 20.0 | 535. |
| LXS-8-15-OV-R** | 45.0 | 41.0 | 36.0 | 30.0 | 580. |

**Built-in continuously adjustable overvoltage protection crowbars output
when trip level is exceeded. Included on all LXS-7, LXS-8 models.

GUARANTEED LX SERIES. DELIVERY.

20 VOLTS $\pm 5\%$ SINGLE OUTPUT

| MODEL | MAX. AMPS AT AMBIENT OF: | | | | PRICE |
|-----------------|--------------------------|------|------|------|--------|
| | 40°C | 50°C | 60°C | 71°C | |
| LXS-CC-20-R | 7.7 | 7.2 | 6.5 | 4.4 | \$210. |
| LXS-D-20-R | 11.5 | 10.2 | 8.6 | 6.8 | 260. |
| LXS-E-20-R | 15.0 | 13.0 | 10.5 | 7.0 | 320. |
| LXS-EE-20-R | 22.0 | 18.5 | 14.5 | 10.0 | 420. |
| LXS-7-20-OV-R** | 28.0 | 25.0 | 20.5 | 15.5 | 535. |
| LXS-8-20-OV-R** | 32.0 | 29.0 | 25.0 | 17.0 | 580. |

24 VOLTS $\pm 5\%$ SINGLE OUTPUT

| MODEL | MAX. AMPS AT AMBIENT OF: | | | | PRICE |
|-----------------|--------------------------|------|------|------|--------|
| | 40°C | 50°C | 60°C | 71°C | |
| LXS-CC-24-R | 6.8 | 6.4 | 5.7 | 4.4 | \$210. |
| LXS-D-24-R | 10.0 | 8.8 | 7.5 | 6.0 | 260. |
| LXS-E-24-R | 13.0 | 11.0 | 9.5 | 6.0 | 320. |
| LXS-EE-24-R | 19.0 | 16.5 | 13.0 | 9.5 | 420. |
| LXS-7-24-OV-R** | 25.0 | 22.0 | 18.0 | 14.0 | 535. |
| LXS-8-24-OV-R** | 30.0 | 27.0 | 23.5 | 17.0 | 580. |

28 VOLTS $\pm 5\%$ SINGLE OUTPUT

| MODEL | MAX. AMPS AT AMBIENT OF: | | | | PRICE |
|-----------------|--------------------------|------|------|------|--------|
| | 40°C | 50°C | 60°C | 71°C | |
| LXS-CC-28-R | 6.0 | 5.6 | 5.0 | 4.3 | \$210. |
| LXS-D-28-R | 9.0 | 8.0 | 6.8 | 5.5 | 260. |
| LXS-E-28-R | 11.0 | 10.0 | 8.5 | 5.5 | 320. |
| LXS-EE-28-R | 17.0 | 15.0 | 12.0 | 9.0 | 420. |
| LXS-7-28-OV-R** | 22.0 | 19.5 | 16.0 | 12.5 | 535. |
| LXS-8-28-OV-R** | 28.0 | 25.5 | 22.5 | 17.0 | 580. |

± 15 TO ± 12 VOLTS DUAL OUTPUT⁽¹⁾

| MODEL | ADJ. VOLT. RANGE VDC | MAX. AMPS AT AMBIENT OF: | | | | PRICE |
|--------------|----------------------|--------------------------|-------|-------|-------|--------|
| | | 40°C | 50°C | 60°C | 71°C | |
| LXD-3-152-R | ± 15 | 0.400 | 0.370 | 0.340 | 0.300 | \$ 90. |
| | to ± 12 | 0.400 | 0.370 | 0.340 | 0.300 | |
| LXD-A-152-R | ± 15 | 1.0 | 1.0 | 0.9 | 0.7 | 130. |
| | to ± 12 | 0.8 | 0.8 | 0.7 | 0.6 | |
| LXD-B-152-R | ± 15 | 1.6 | 1.4 | 1.2 | 0.7 | 160. |
| | to ± 12 | 1.4 | 1.3 | 1.1 | 0.6 | |
| LXD-C-152-R | ± 15 | 2.5 | 2.3 | 1.9 | 1.5 | 170. |
| | to ± 12 | 2.0 | 1.8 | 1.5 | 1.2 | |
| LXD-CC-152-R | ± 15 | 4.0 | 3.7 | 3.2 | 2.4 | 255. |
| | to ± 12 | 3.0 | 2.7 | 2.3 | 1.8 | |
| LXD-D-152-R | ± 15 | 6.2 | 5.6 | 4.9 | 4.0 | 300. |
| | to ± 12 | 4.5 | 4.1 | 3.7 | 3.0 | |
| LXD-EE-152-R | ± 15 | 12.5 | 11.0 | 9.0 | 7.0 | 455. |
| | to ± 12 | 10.0 | 9.0 | 7.8 | 6.0 | |

Overvoltage protector accessory available for all models without built-in overvoltage protection.

(1) ± 15 to ± 12 and ± 6 to ± 3 volts are each dual tracking outputs; dual outputs can be connected in series for 30-24 volts and 12-6 volts, respectively.

± 6 TO ± 3 VOLTS DUAL OUTPUT⁽¹⁾

| MODEL | ADJ. VOLT. RANGE VDC | MAX. AMPS AT AMBIENT OF: | | | | PRICE |
|-------------|----------------------|--------------------------|------|------|------|--------|
| | | 40°C | 50°C | 60°C | 71°C | |
| LXD-B-062-R | ± 6 | 2.7 | 2.4 | 1.9 | 1.4 | \$170. |
| | to ± 3 | 2.1 | 2.0 | 1.6 | 1.2 | |
| LXD-C-062-R | ± 6 | 3.5 | 3.3 | 2.7 | 1.7 | 180. |
| | to ± 3 | 2.6 | 2.4 | 1.9 | 1.3 | |

5 VOLTS $\pm 5\%$, ± 15 TO ± 12 VOLTS TRIPLE OUTPUT

| MODEL | ADJ. VOLT. RANGE VDC | MAX. AMPS AT AMBIENT OF: | | | | PRICE |
|--------------|----------------------|--------------------------|------|------|------|--------|
| | | 40°C | 50°C | 60°C | 71°C | |
| LXT-D-5152-R | $5 \pm 5\%$ * | 12.0 | 11.5 | 11.0 | 9.5 | \$395. |
| | ± 15 | 3.1 | 2.7 | 2.2 | 1.7 | |
| | to ± 12 | 2.3 | 2.0 | 1.7 | 1.3 | |
| | | | | | | |

*5 volt output has fixed overvoltage protection at 6.8V $\pm 10\%$. ± 15 to ± 12 output is dual tracking output.

NEW LX-7 DESIGNED TO MEET MIL ENVIRONMENTAL SPECIFICATIONS.



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Servicing a high-technology product

means that engineering must provide a design capability that minimizes later problems with the customer.

As a company whose goal is to do complete servicing for every product we design, we're acutely aware of a problem that nearly all electronics manufacturers face: How to design products that won't give the end user trouble—or, if they do, that can be serviced easily.

Our engineering staff aims to do it by meeting the following challenges:

- Providing full product characterization.
- Using long-life components.
- Realizing that compromise designs are possible.
- Providing complete documentation.

The push beyond minimum/maximum specs

Any company that is going to service its own product must determine the full characterization of the products. It must find out everything its products can do—not just the minimum and maximum specs, but everything. That will also provide a complete list of applications for potential customers.

Here's an example of what can happen if you don't push for full product characterization: We had developed a fine product and introduced it to the market before most of the competition. We found a good demand for the unit, and there were many questions about its performance.

A full characterization of it had not been done, so we had to do a lot of tests to answer the questions of potential users. We decided from then on that all our products should go through a prototype stage, making sure that we don't skip any steps in the full characterization. It was an expensive process, but in the long run it was a more efficient technique than doing a number of spot tests.

It also saved the potential user the job of doing a lot of the testing himself. Then, too, if you characterize the product, you may find out that there are some parts of the item that

can be improved with only a minor change. That'll cut down on your servicing time later.

Figuring the component tradeoffs

Another major challenge that always comes up in any major engineering product is the component question. Since the project manager wants to produce a device that will be sold for several years, should he use established components or newer components?

The problem in using older components—transistors vs integrated circuits, say—is that with the older components, the design may be easier and more quickly done but the cost may be higher. It costs more, for example, to use many transistors instead of a few ICs.

The engineer must look into the future and figure which costs will come down; he must equate material costs against the material plus associated labor for the older components.

To solve the component challenge, it's important for the manager to emphasize future use. He has to make sure that engineering personnel are informed daily of presently available components and at the same time make sure that they have a feeling as to what the future projects.

If the engineer designs something that contains a single-sourced component, part of his future projection should consider whether there'll be multiple sources in the future.

Specs are made to be broken

Any engineer who works for a high-technology company must realize that some specs are written to be compromised. With any particular product design, there's always an initial specification that engineers are designing against. That spec is basically a guide. It's very easy for an engineer to say: "Spec 'A' is what I need, and I'm going to keep butting my head against the wall until I achieve it."

The other, more practical alternative to this is for engineering management to present the philosophy that the specs are a goal and that in

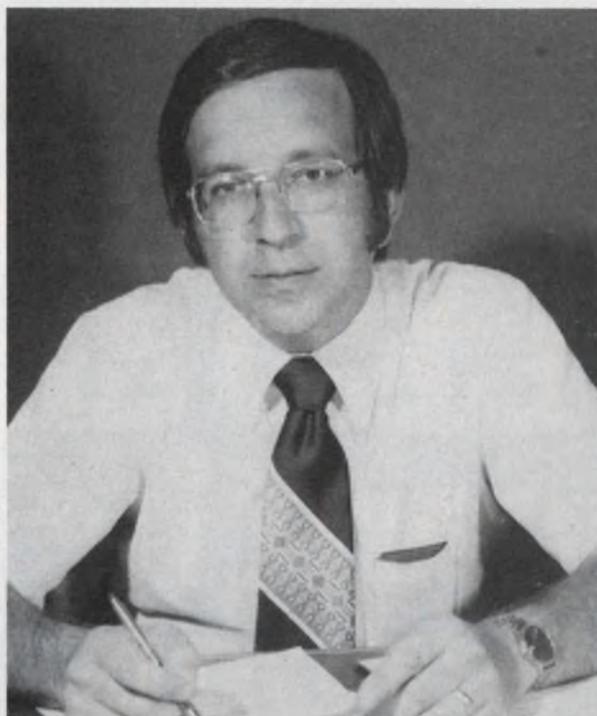
Jack Alford and DDC

During the 8-1/2 years that Jack Alford has been with ILC Data Device Corp., he has held positions of Product Engineer, Director of Engineering, and Director of Micro-electronics. He is now Director of Product Planning.

Prior to joining ILC DDC, he was with Telemet Company as Product Line Manager for supervisory control instrumentation and sub-systems. Before that he was Product Engineer for Loral Electronics Corp. and for Sperry Gyroscope in countermeasures systems.

ILC DDC is a wholly owned subsidiary of ILC Industries, which is owned by Rapid American. The company has been in business for approximately 10 years and is presently located in a 50,000 sq. ft. building in Bohemia, N.Y. Present employment is over 300 people.

DDC originally manufactured operational amplifiers and subsequently became a major factor in the synchro conversion and video digitizing area. The company also has been growing at a rapid rate in producing thick film hybrid circuits.



the real world those goals can be compromised. It's important for the engineer to realize at that point that if he runs up against this brick wall, he should find out where he's running against it and come up with a good alternative.

Tradeoffs are always possible. The thing to remember is that this is real life, and real-life people are willing to make compromises.

Making sure the product can be produced

In the initial stages of any company a certain number of products are released before the design is completely debugged and before production has been set up to assure producibility. One management problem is to make sure that appropriate systems are set up to ensure that all necessary documentation and design reviews are complete. This is particularly important when the company services its own products.

After growing from a small company to a larger one, we initiated a unilateral engineering review of all new-product designs. As we progressed, we found that the best method was a joint review of design among management, design and manufacturing personnel. This helped to uncover some more of the difficulties that come up in any normal design, and it eliminated the bulk of them. Engineering must always realize that its job isn't done until manufacturing is capable of producing the design with minimum assistance from engineering.

And two obvious but important points before your company can even begin to meet its servicing goal: The product must be designed on time at a reasonable cost, and it must be a product that will sell. Unless it meets these essentials, the company won't stay in business long.

One of the most important jobs any electronics manufacturer faces is to make maximum use of the engineer's time. Many engineers tend to optimize their designs beyond the point that's necessary. The moment the engineer has met his specs is the moment he should stop designing and determine how much time he has left before the deadline. If he has enough to improve the product, fine. But if he doesn't, he must realize that the timing for introduction of a new product takes precedence over improving the product beyond its specs.

If the engineer keeps stretching the schedule for an additional three to six months, the competition may have enough time to produce an equivalent product. Engineers must interface with marketing enough to know that if there isn't a well-defined time factor for product introduction, the whole product ballgame is lost.

And a company might as well stay away from the ballpark altogether if its engineers are hooked on designing only products that are technological milestones. Engineering managers should make it clear to their staffs that the idea for new products is only as good as the market will accept. ■■

20 million LED digits can't be wrong.

Litronix DL-707 0.3" high digit has same superior features as DL-747 below.

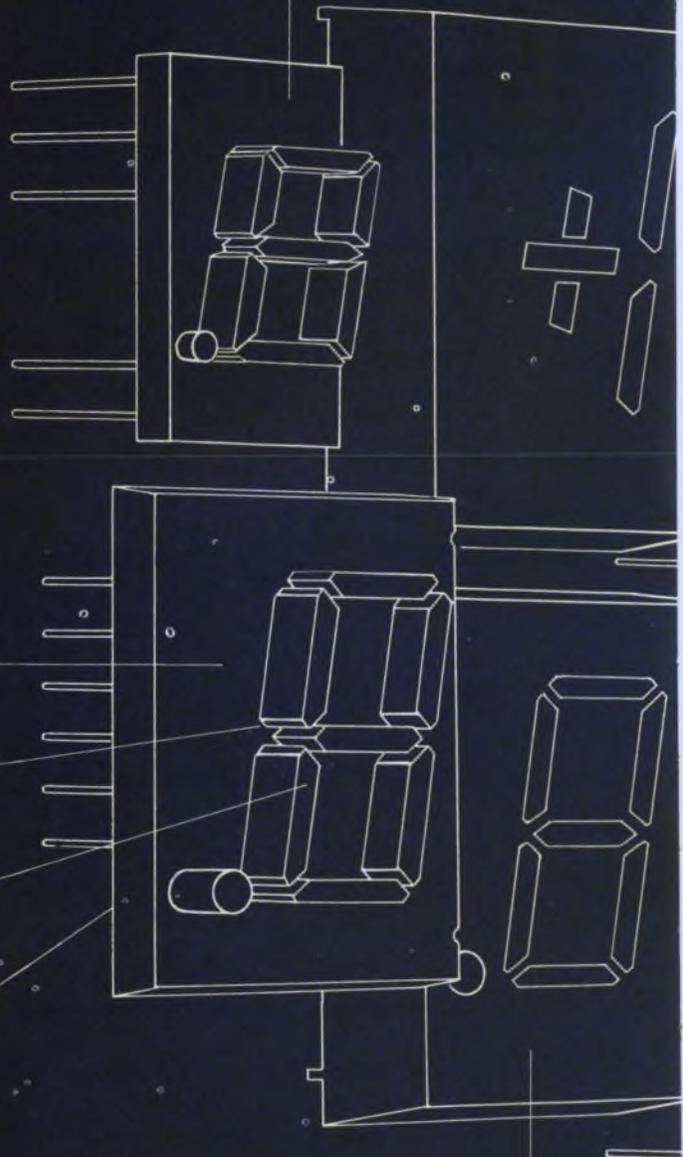
Litronix DL-747 has 0.6" high digits 44% larger in area than digits from any other major supplier.

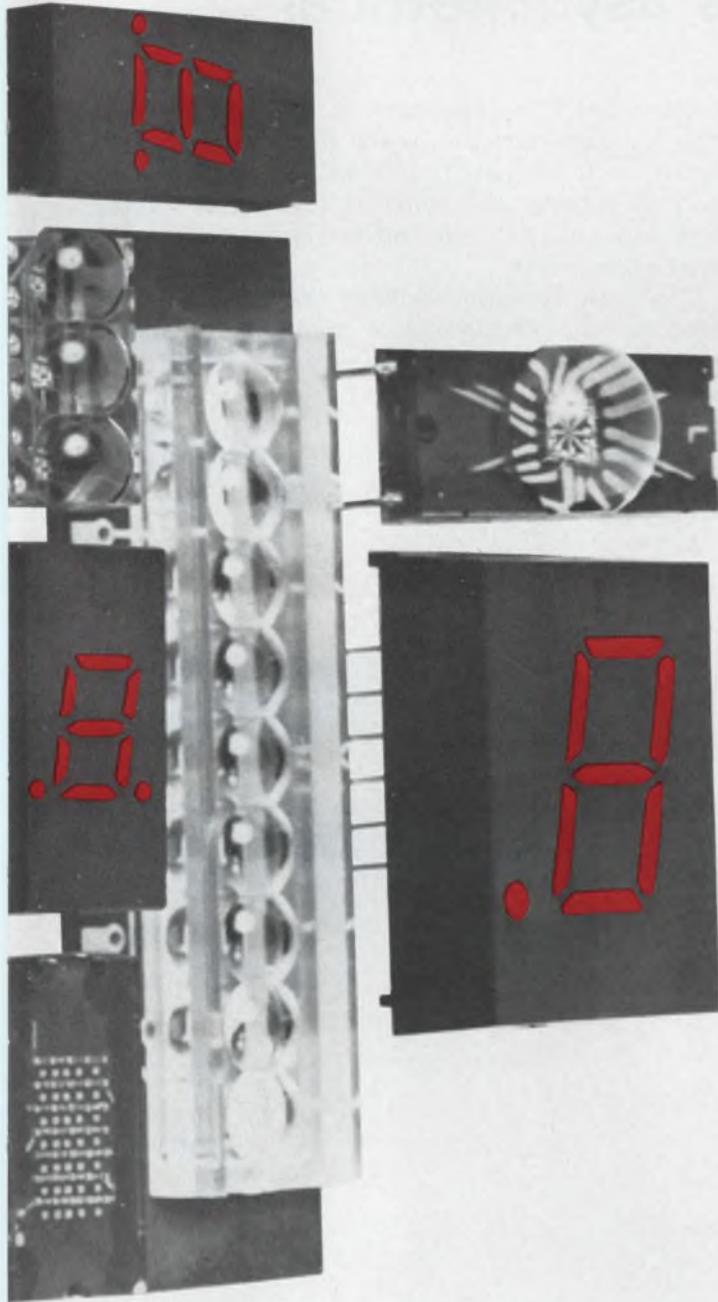
Mitered corners increase eye appeal.

Light pipe spreads illumination evenly.

Mounted on DIP with standard pin spacing.

DL-747 with cap in place over light pipes.





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For openers, the Litronix product line is so broad that you can get exactly what you need. Sizes from 0.1" to 0.6". Colors like red and orange today, with yellow and green in volume production soon. Common anode or common cathode. And polarity overflow digits in many sizes.

Perhaps even more important, you're assured of a consistent, reliable product. Litronix doesn't buy any high-technology parts from others. We make everything in-house—starting with the basic GaAsP materials. The result is that we have better overall control: high brightness, proven reliability and assured delivery.

Litronix has two plants in the U.S. and three more overseas to keep the LEDs flowing. Because our volume is large, our prices are very, very competitive.

The drawing at left of our big 0.6"-high DL-747 shows some of the superior characteristics of the popular Litronix 700 Series displays.

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Counter has symmetrical output though the input signal is asymmetrical

The divide-by-five synchronous counter shown in the figure provides a 50%-duty-cycle output with an asynchronous input. This feature is unusual, since most odd pulse-counter circuits require a symmetrical input to obtain a symmetrical output.

The asymmetry that the circuit can handle is limited only by the minimum pulse width that is determined by the propagation delay time of FF_3 and the data setup time required by FF_1 . The manufacturer's maximum specifications of 45 ns and 20 ns, respectively, for these time intervals yield a minimum clock input pulse width, t_w , of 65 ns for reliable operation.

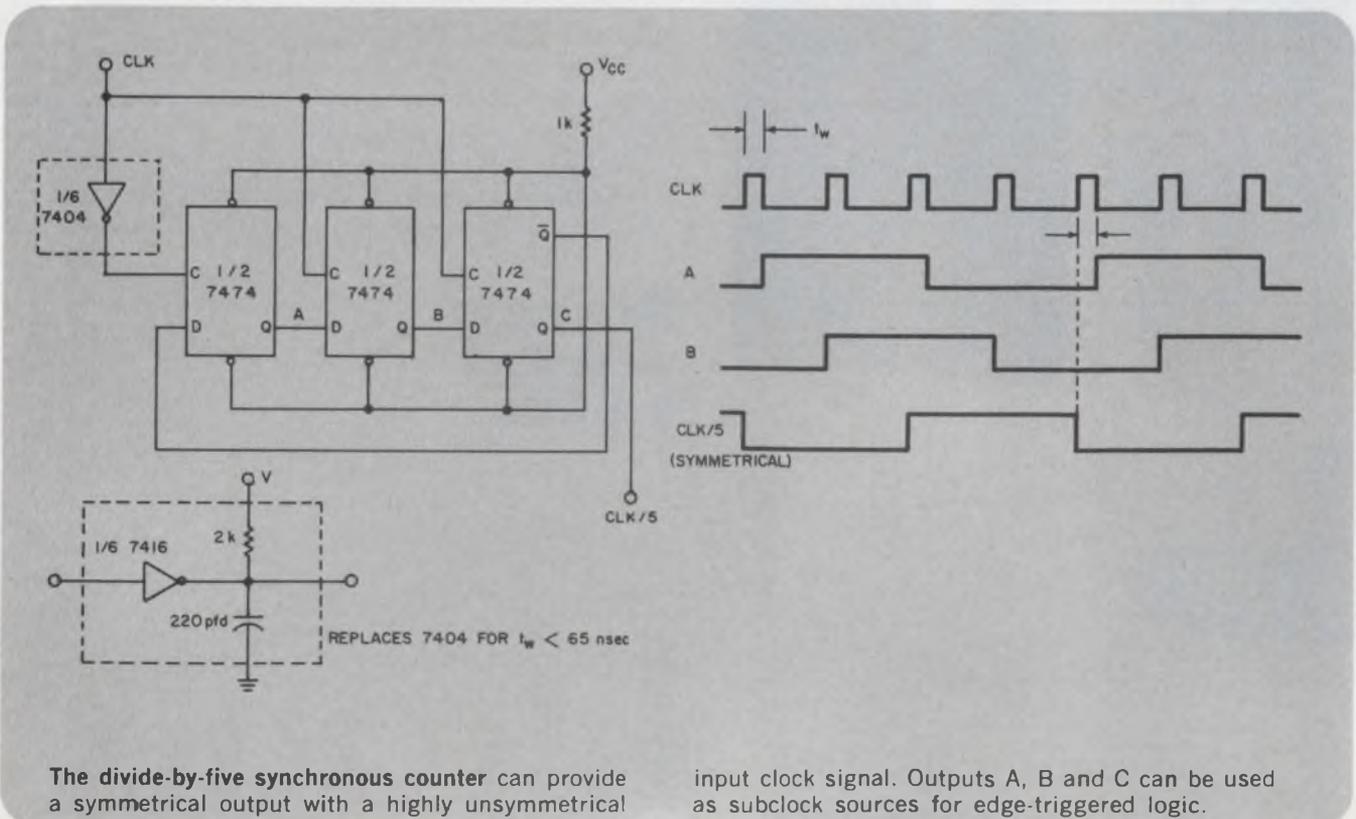
Pulse widths between 30 and 65 ns can be ac-

commodated by replacement of the 7404 inverter with an open-collector buffer like the 7416 (see dotted box). The RC time constant of the collector load reduces the slope of the positive-going clock input to FF_1 , and effectively stretches the input pulse width.

If a logic designer follows the example provided by the divide-by-five counter, he should easily be able to modify the circuit for different moduli.

M. Barry Greenberg, Project Engineer, Systems Div., GB Instruments, 2030 Coolidge St., Hollywood, Fla. 33020.

CIRCLE No. 311

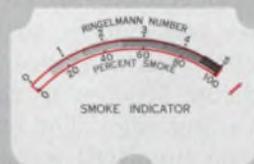
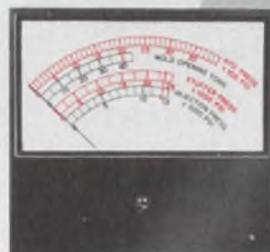
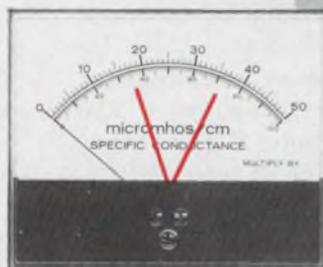
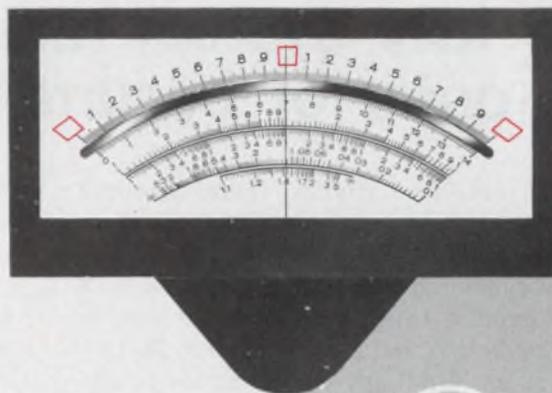


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INFORMATION RETRIEVAL NUMBER 63

Simple tally system ranks events in the order of occurrence

Behavioral-science studies, entertainment games and special logic-design problems need an order-of-occurrence circuit to rank a sequence of events with nanosecond unambiguity. The circuit in the figure handles three events, and it can be expanded to cover any number of events. Merely add event-register flip-flops, extra inputs to the Input-OR circuit and the expansion of the rank shift registers and displays.

To start an event cycle, all rank shift-register stages except the first (A_1 , B_1 , C_1) are reset. The first-rank flip-flops are set to logic ONEs. Input pulses for each event can be derived from switches or other transducers, such as photocells.

An event pulse triggers its corresponding event-register flip-flop to the ONE state on the pulse's trailing edge. The event pulse also passes through the input-OR circuit. But the delay inherent in several serially connected gates allows the Event-Register flip-flop to block its associated rank register from shifting.

Thus if event A occurs first, the first-rank flip-

flop of A remains in its present ONE-state, but the second-rank flip-flops of the B and C rank registers go to ONE.

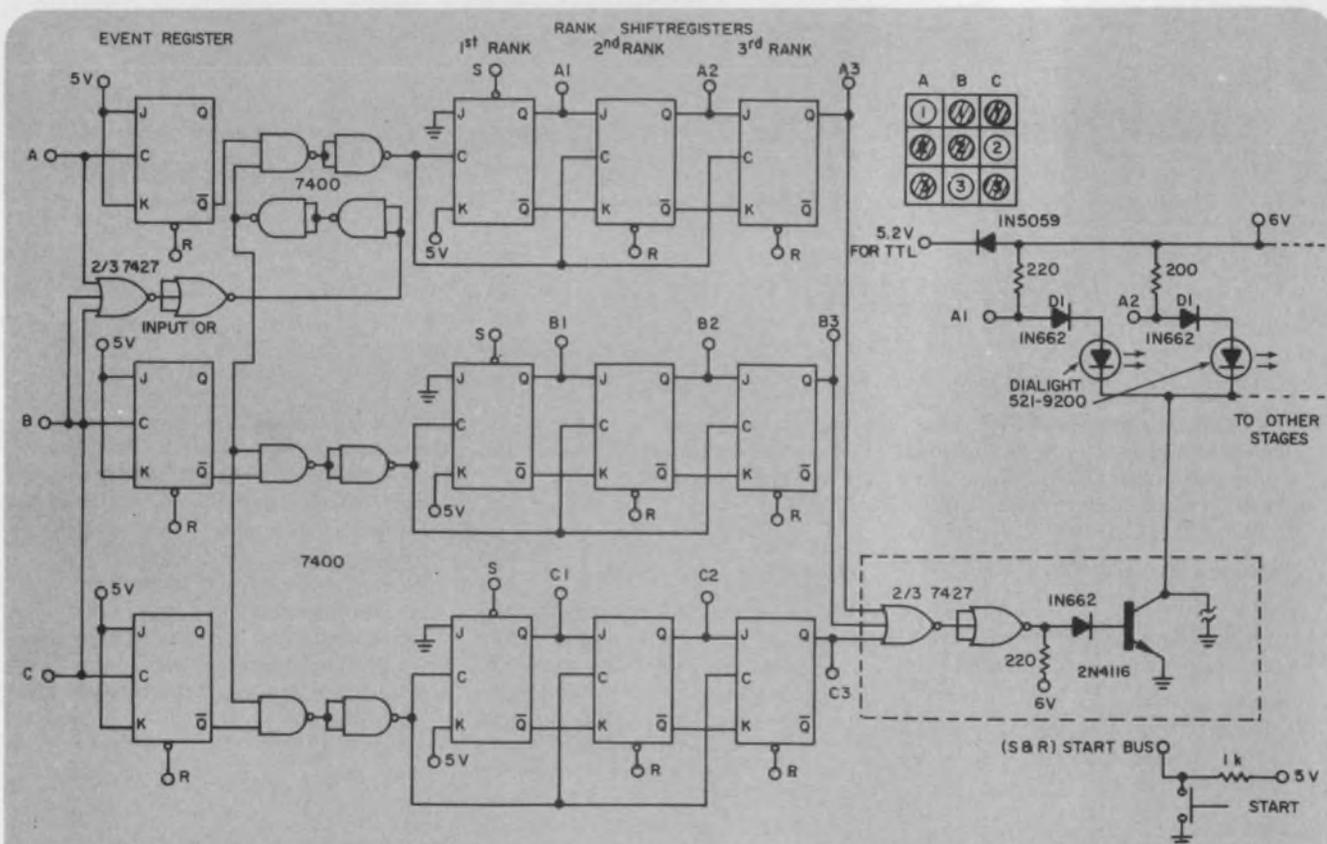
The next event pulse, say C, sets event-register C so the C second-rank flip-flop remains high, but the B-rank register shifts a ONE to the third rank. Thus A is ranked first, C second, and B is third.

The rank shift register outputs are A_1 through A_3 , B_1 through B_3 , etc. Each output connects to a corresponding lamp driver. The figure shows typical lamp drivers for A_1 and A_2 . The bulbs are arranged on a tally scoreboard as shown.

If the display should light only at the "end-of-play," add the section shown within the dotted box. The Q outputs of all the third-rank flip-flops are ORed to enable display of the tally only at the end-of-play.

Sujit R. Kumar and G. Dutta, Computer Centre, Japavpur University, Calcutta 700032, India.

CIRCLE No. 312



A tie-breaker circuit ranks the occurrence of event pulses with nanosecond precision. A delay in the input-OR circuit is necessary to avoid a race condition and allow the event pulses to toggle the

event-register flip-flops before the pulses can clock the rank shift registers. The delay is obtained inexpensively by use of cascaded "left-over" gates.

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| ± 15 | .100 | .05 | .05 | 1 | 3.5x2.5x1.38 | 55.00 | DB15-10 |
| ± 15 | .150 | .05 | .05 | 1 | 3.5x2.5x1.38 | 65.00 | DB15-15 |
| ± 15 | .200 | .05 | .05 | 1 | 3.5x2.5x1.38 | 75.00 | DB15-20 |
| ± 15 | .300 | .05 | .05 | 1 | 3.5x2.5x1.63 | 105.00 | DB15-30 |
| ± 15 | .350 | .05 | .05 | 1 | 3.5x2.5x1.63 | 110.00 | DB15-35 |
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INFORMATION RETRIEVAL NUMBER 64

Inexpensive AM modulator replaces clipping type and gives less distortion

A simple amplitude modulator provides a cleaner output signal than conventional clipping modulators. The improved circuit removes or adds a signal to the middle portion of the carrier sine wave instead of clipping the height of a square wave. Using only three op amps and some passive components, the circuit requires no initial adjustment, is temperature-stable, has a wide modulation-frequency range and needs no filtering—and hence no returning if the carrier frequency changes.

The input to A_1 (Fig. 1) is the sum of the carrier and modulation signals. Amplifier A_1 clips the composite input to produce the output shown in Fig. 2a. Amplifier A_2 produces a similar signal (Fig. 2b), except that the carrier phase is inverted. The output of A_2 is then inverted and summed with the output of A_1 in the output amplifier A_3 .

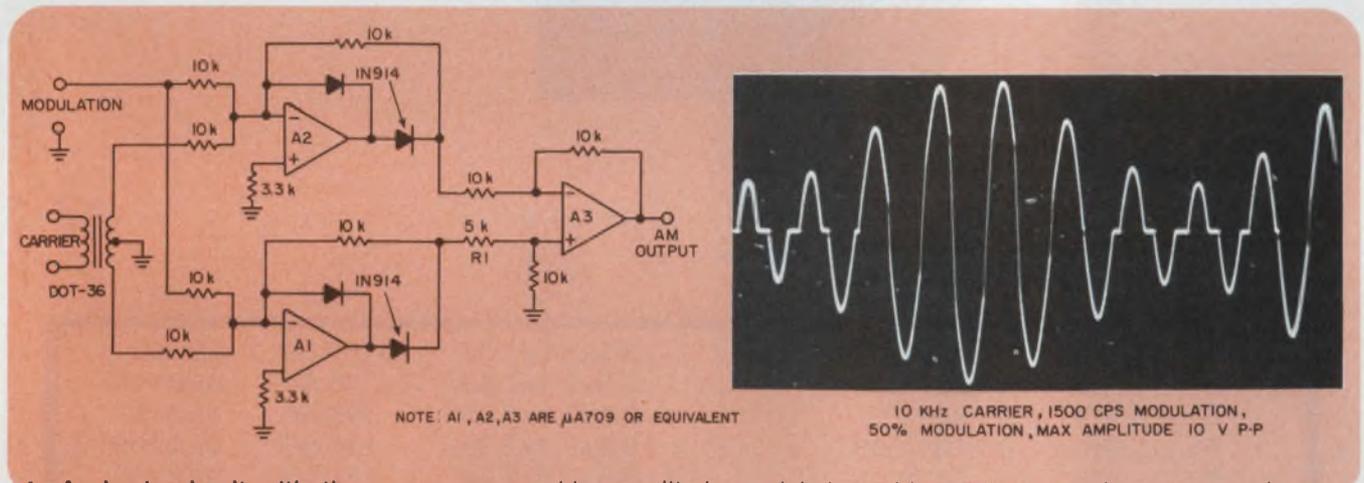
Diode feedback paths around A_1 and A_2 provide sharp clipping, even with small input amplitudes. However, the clipper output impedances are nonlinear, varying from near zero to approximately 10 k Ω when a negative-going out-

put is being clipped. This means that Fig. 2b doesn't accurately represent the output of clipper A_2 . The waveform at that point will, because of resistive division and mixing, contain a small portion of the waveform of Fig. 2a. To compensate for the output-impedance variation, resistor R_1 at A_3 's noninverting input is made 5 k Ω .

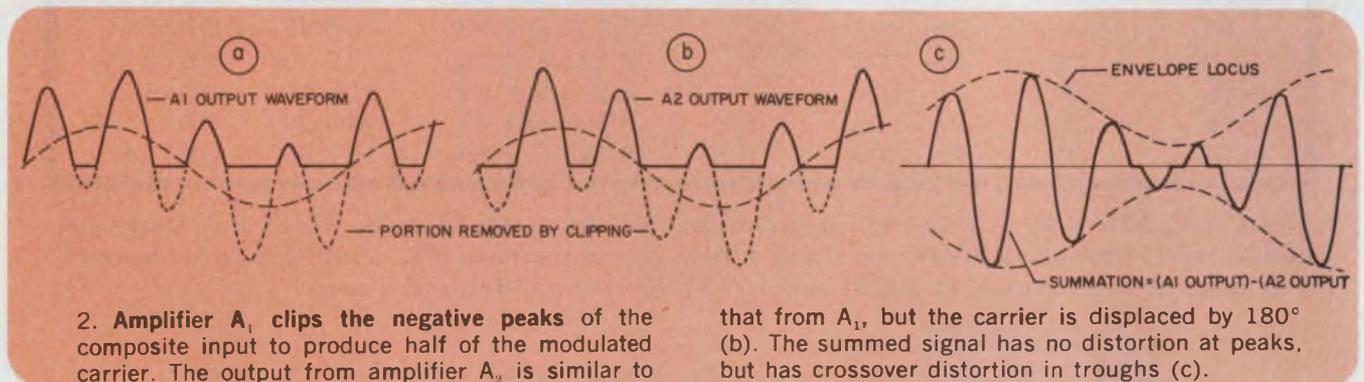
Distortion in the net output signal (Fig. 2c) appears only in the valleys of the modulation envelope. This distortion depends on the modulation percentage and is most objectionable near minimum output. However, the circuit doesn't distort the peaks of the modulated waveform. Therefore peak detectors will respond to the signal of Fig. 2c as if it were a distortionless AM signal.

The circuit was built with Type 709 op amps and performs well with carrier frequencies up to 30 kHz. At higher frequencies the rate-limiting of the op amps restricts the output amplitude.

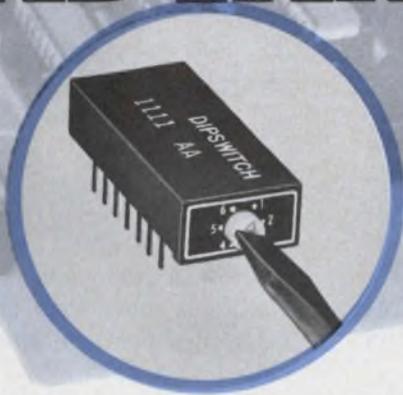
Penn Clower, Design Engineer, Charles Stark Draper Laboratory, 68 Albany St., Cambridge, Mass. 02139 CIRCLE No. 313



1. A simple circuit with three op amps provides amplitude modulation without distorting the output peaks.



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Comparator detects volts in narrow window and accurately nulls to reference level

A very accurate window-comparator circuit can be made by insertion of an integrator into the feedback loop of an op amp. The integrator servos out the effects of offset and drift voltages of both amplifiers.

In the circuit (see figure), this servo action keeps point e_s within a few hundred microvolts of e_A , and the value of e_R can range over ± 10 V, independent of the window size ($2\Delta V$). The window size is determined by V_z , which is the maximum voltage output of A_2 . At null within the window region $e_A \approx e_s$, the two inputs of A_1 behave as if they were connected to make e_R sink the current i_s . Thus:

$$e_R/R_3 = (E_i/R_1) \pm (V_z/R_2).$$

At the edges of the window range, V_z is approximately ± 13 V when an LM301 amplifier is used for A_2 .

If we make $R_1 = R_3$ (though R_1 need not equal R_3) for $E_i = e_R \pm \Delta V$, then $0 = (\pm \Delta V/R_1) \pm 13/R_2$, or $\Delta V = \pm 13 R_1/R_2 = 1.3$ mV, when $R_1 = 100 \Omega$ and $R_2 = 1 \text{ M}\Omega$.

For a value of E_i that falls outside ΔV , amplifier A_1 saturates, and its plus and minus outputs are converted into a ZERO logic level via Q_1 , Q_2 and Q_3 . Voltage ΔV is entirely a function of A_2 's maximum output stability. However, the stability of the window is improved considerably if A_2 is bounded with zeners.

The circuits containing Q_1 , Q_2 and Q_3 convert the output of A_1 to a unipolar function for logic interface.

The long-term stability of this circuit is excellent, since amplifier drift is constantly corrected relative to the reference voltage. Assume the window reference voltage e_R is at zero. The offset voltage of A_1 is represented as e_A and the servo voltage as e_s .

$$\text{Then } A_1 (\text{out}) = -A_1 (e_A - e_s),$$

$$A_2 (\text{out}) = -A_2 (-A_1) (e_A - e_s),$$

if we neglect e_B

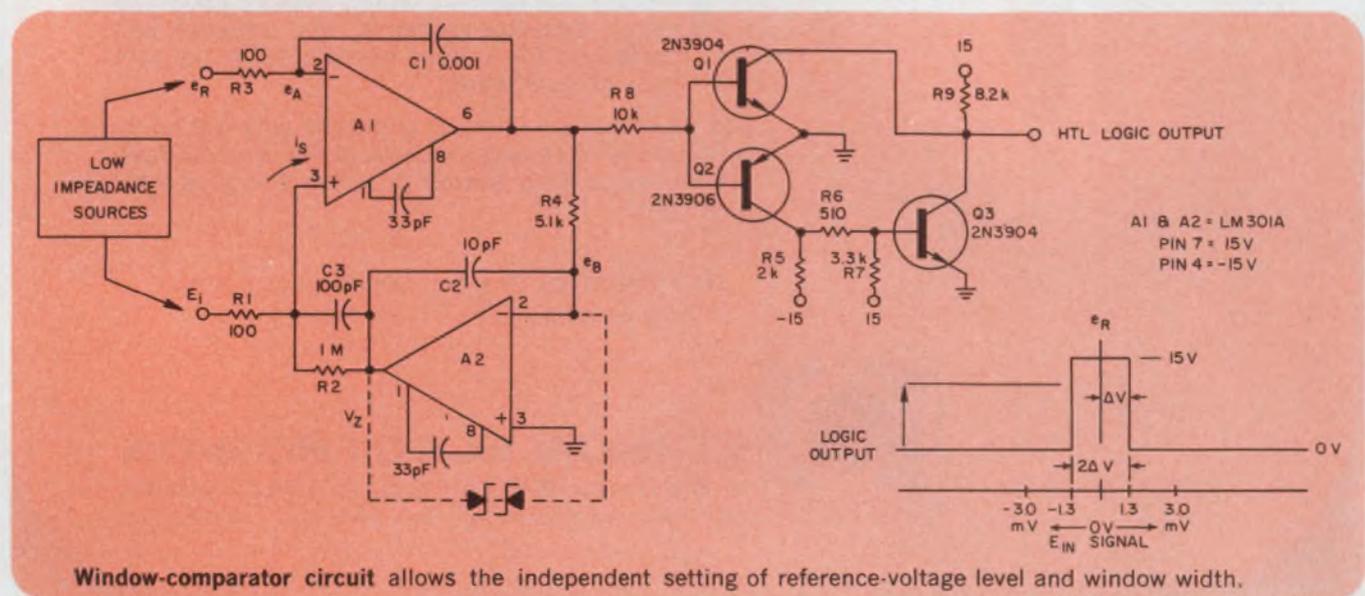
$$e_s = A_1 A_2 e_A - A_1 A_2 e_s,$$

$$\text{and } e_s = \frac{A_1 A_2 e_A}{A_1 A_2 + 1} \approx e_A.$$

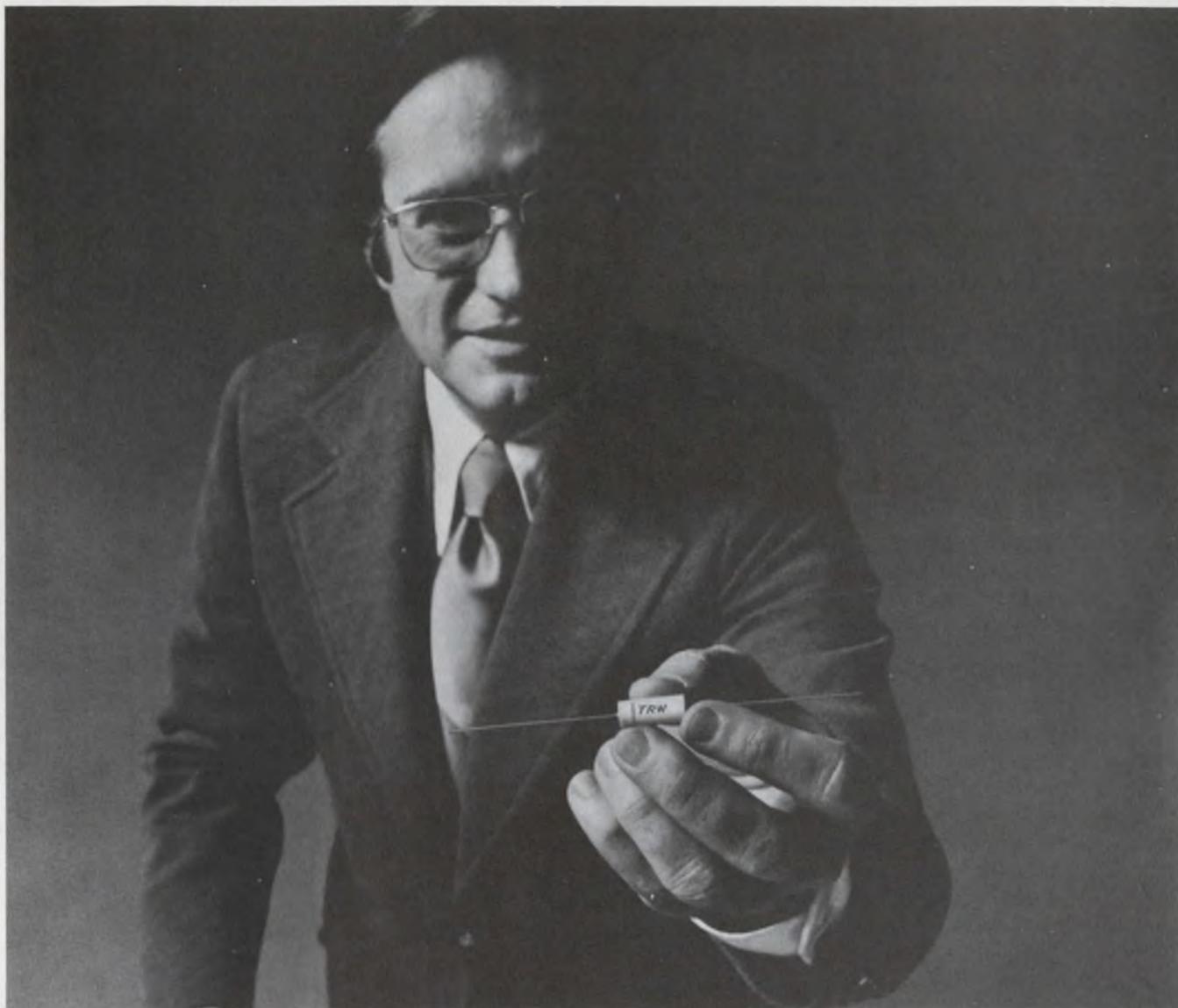
Similarly the offset voltage, e_B , of amplifier A_2 is $e_s \approx e_B/G_1$, where G_1 is the closed-loop gain of A_2 . Thus e_A will equal e_s within 0.004% for an A_2 open-loop gain of 25,000, which is the minimum performance for a 301A op amp. The closed-loop gain of A_2 equals unity.

The 0.004% unbalance results in less than 1 $\mu\text{V}/^\circ\text{C}$ integrator drift. This compares with a typical drift of 60 $\mu\text{V}/^\circ\text{C}$ in conventional window comparators, which usually contend with the drift of two amplifiers. Obviously an amplifier with high open-loop gain will give better performance, but a gain of 10,000 is generally more than adequate.

Speed is not of prime consideration in many comparator circuits—especially in low-level window detection, where long-term stability and low drift are the main criteria. In the servo-nulled circuit, speed is predominantly a function of the op-amp slew rate for slowly changing input signals. However, the effect of C_1 , C_3 , R_2 and R_3



Window-comparator circuit allows the independent setting of reference-voltage level and window width.



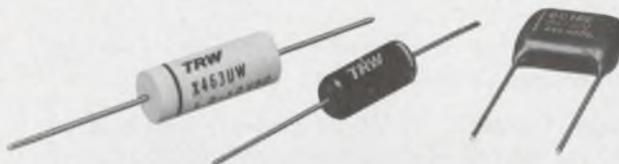
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INFORMATION RETRIEVAL NUMBER 66

IDEAS FOR DESIGN

become dominant as they increase in value.

Of course, a fast FET amp for A_1 and a fast bipolar for A_2 could improve the response time and allow the use of larger values for R_1 and R_2 . For slowly changing input signals, the response delay at the logic output is about $5 \mu\text{s}$. For a fast-step input or a high frequency superimposed

on dc, C_3 acts a momentary short across R_2 to increase the window width temporarily. This short lasts as long as $500 \mu\text{s}$ for the values shown and it eliminates input voltage transitions in the output during this period—a form of high-frequency and transient-noise filtering.

K. R. Johnson, Canaco Inc., 5635 Fisher Lane, Rockville, Md. 20852. CIRCLE No. 314

Counter resets itself reliably with one additional flip-flop

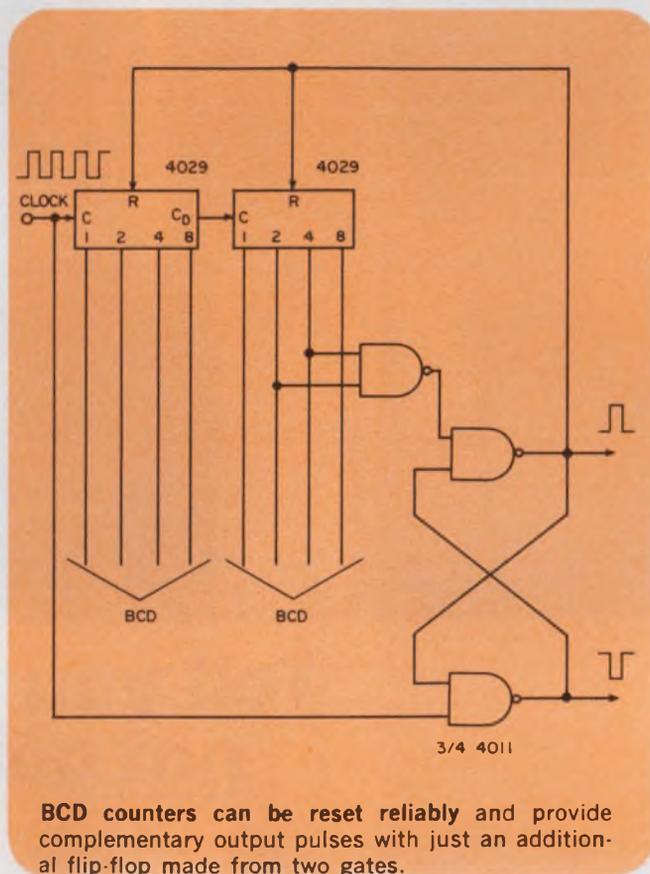
One additional set-reset flip-flop allows you to reset reliably a BCD counter that divides a pulse train by a preset number. The maximum count desired is detected with a decoder gate. However, if the output of the decode gate is used directly to reset the counters, race problems can occur.

Without this circuit, as soon as 1 bit of the counter returns to ZERO, the decode condition is lost. A very short reset pulse results, which might not reset all the counter bits properly. The difficulty can be overcome by addition of a flip-flop circuit to the reset path.

As an example, the circuit in the figure uses CMOS 4029 counters to achieve division by 60 with BCD outputs. Both the clock and reset inputs for the 4029 operate on a rising edge. The count of 60 is decoded by a NAND gate to set the flip-flop, which in turn resets the counters. The flip-flop, however, does not reset until the trailing edge of the clock. This results in a square reset-pulse output, half the width of the clock pulse train. The reset pulse, or its complement, is available as a carry-out.

John Budlong, Design Engineer, Western Research & Development Ltd., Lab #3, 1313-44th Ave. N. E., Calgary, Alberta, Canada T2E-665.

CIRCLE No. 315



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Equipment shows errors in sonar communication

A prototype of error-detection and correction equipment for teleprinter-to-teleprinter underwater sonic communication has been developed by Marconi Communication Systems Ltd., Chelmsford, England. The equipment—Sonar 2010—was designed in collaboration with the Royal Navy Admiralty Underwater Weapons Establishment.

Initial studies at sea investigated techniques for solution of the problems in this type of data transmission. Such problems include fading, multipath distortions, noise bursts and Doppler shifts that are caused by the relative motion between communicating vessels.

During sea trials of the proto-

type equipment, the Marconi underwater telephone Sonar 2008 was used to transmit and receive telegraph signals. The performance target of 98% error-free copy, under a wide range of propagation conditions from shallow water, deep sea and rough sea to calm tropical conditions, was exceeded.

The 2010 system sent traffic at above speech transmission rates. The greatly increased range provided by the 2008 underwater telephone has extended the communications abilities of the submariner. Production, scheduled to start this year, will cover a ship-fitting program of existing and new Royal Navy vessels over a period of years.

Laser method detects IR absorbing gases fast

A laser system for detection of methane and other infrared absorbing gases has a fast response (less than 1 second), high gas-flow rate and a wide range of sensitivities. Unlike flame-based techniques, the laser method measures concentrations of infrared absorbing gases in inert atmospheres and, unlike some solid-state detectors, cannot be affected by the gases.

Developed by International Research & Development Co. Ltd. and A.M.G. Zuurbier Ltd. of Doncaster, England, in collaboration with the Safety in Mines Research Establishment, the system will be used in natural-gas safety surveying, petrochemical safety monitor-

ing or concentration measuring, and high-sensitivity, gas-leakage alarm systems. In its first application, the equipment has been installed in a patrol vehicle for the detection of small quantities of methane; for example, 100 parts in 10^6 .

Operation of the system relies on the absorption of $3.39\text{-}\mu\text{m}$ radiation by the sample gas in a cell located within a laser cavity. Gases and vapors that absorb strongly in the $3.39\text{-}\mu\text{m}$ band include butane, ethane, propane, hexane, heptane, butadiene, dimethylamine, dimethylether, ethylether, ethyl mercaptan, ethylene, ethylene oxide and propylene.

CIRCLE NO. 319

Yagi-array size shrunk by dielectric powder

Yagi and resonant-slot antennas can be reduced in size by use of dielectric coatings, according to researchers at the Royal Military College of Science in England.

Experiments on a three-element Yagi array that was coated with barium titanate powder tuned the antenna, and also lowered the radiation resistance. The radiation back lobe increased, but that was remedied by reduction of the element spacing. The use of fired-ceramic coatings—thinner coatings than have formerly been possible—with relative permittivities of 3000 are expected to reduce antenna lengths more than one-third.

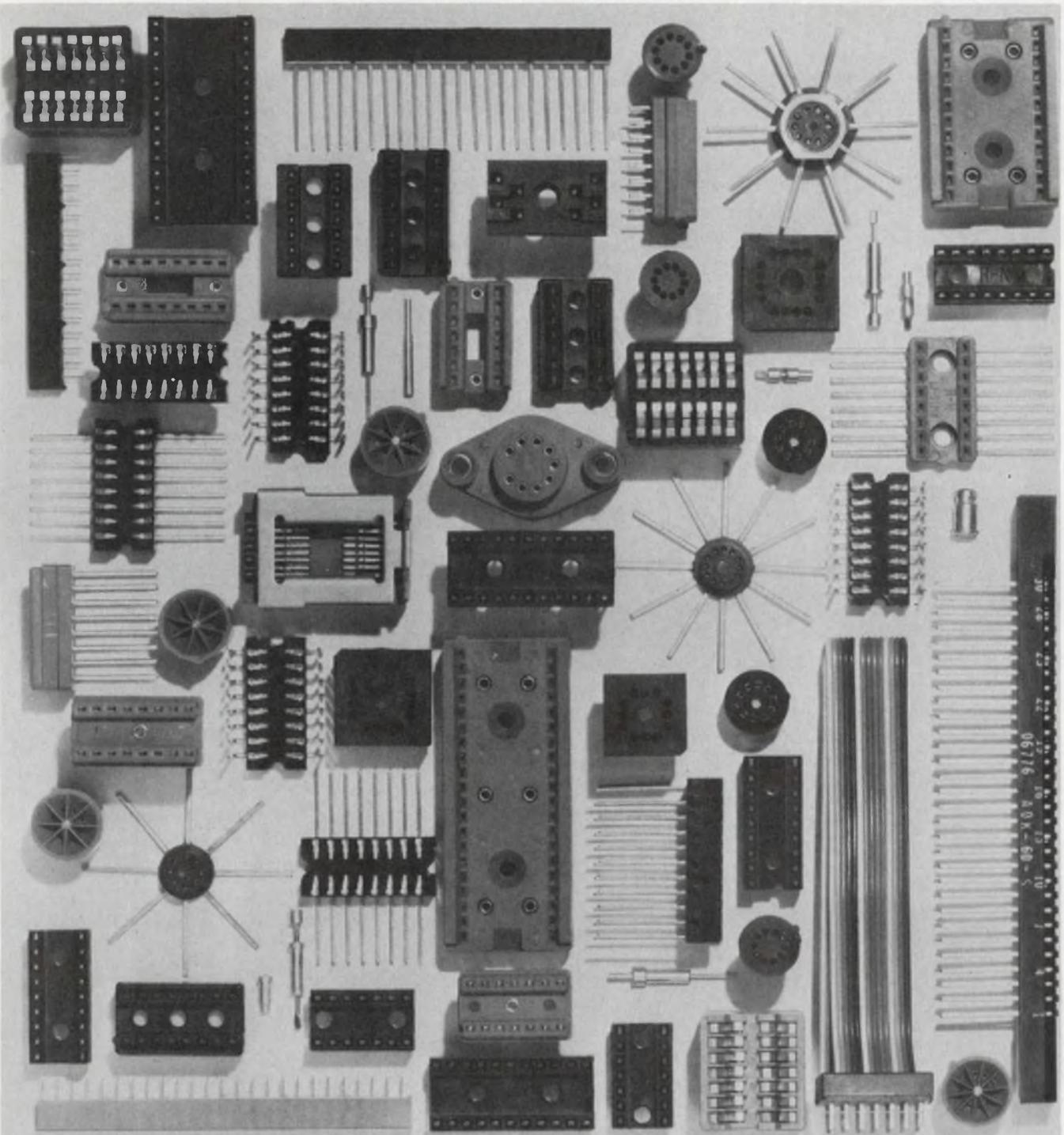
Aircraft slot antennas have been tested with titanium-dioxide powder held on the antenna by a thin plastic skin. At 1 GHz, the length of the slot was reduced by 20% at resonance. There is a loss incurred in the dielectric coating material, typically 2 dB at 100 MHz. However, applications are seen in radio and TV wherever small size is more important than power loss.

Bubble memory built on wafer-thin film

A magnetic-domain memory that stores one million binary digits on a wafer the size of a postage stamp could eventually replace disc and tape memories, according to General Telephone & Electronics in Switzerland.

Still in the prototype stage, the new memory consists of thin film of magnetic material deposited on a garnet crystal. The film, no thicker than a human hair, contains minute cylindrical magnetic domains, known as bubbles, that can be moved at high speed within the material. Bubble movements are controlled by "tracks" of magnetic material placed on the garnet chip.

The presence or absence of bubbles provides the zero/one digital information that can be read by a detection device. Since magnetic-bubble devices are small, economical and reliable, their use in large memories has great potential. A model is being evaluated in a telephone-switching system.



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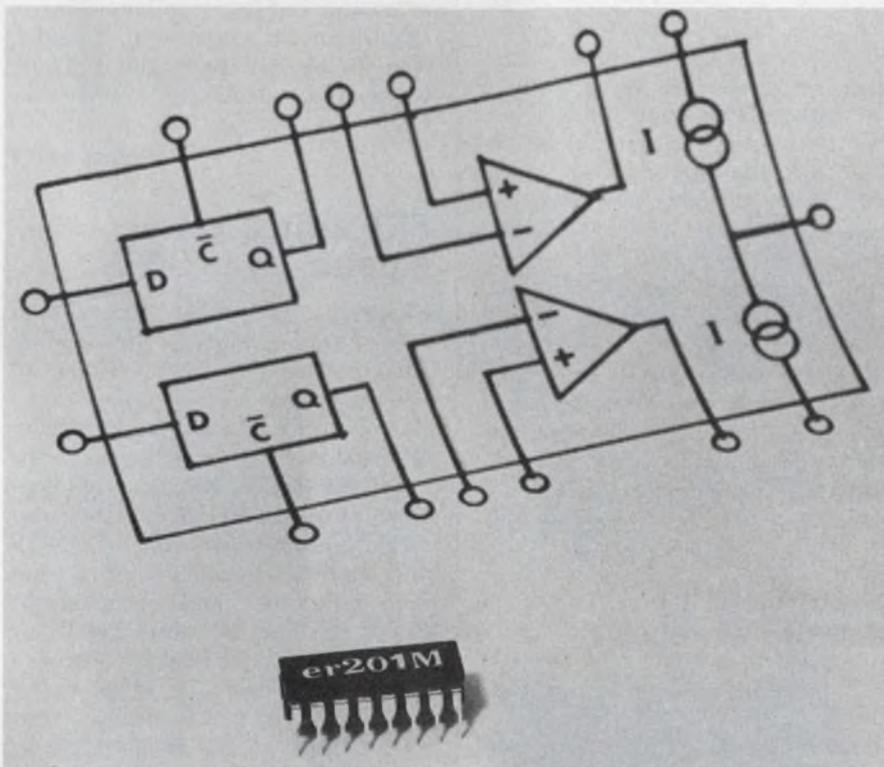
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samplers. All of these sections are externally accessible. Voltages applied to the samplers' inputs form test or reference signals. Analog sampling can be performed at 200-kHz rates, minimum.

When strobed, each sampler corresponds to a high-speed, unity-gain buffer plus JFET output. The sampler has an offset of 30 mV maximum, 10 mV typical, and it requires a bias current of 2 μ A. Sampler cutoff and recovery are obtained at rates of 10 V/ μ s and 50 V/ μ s, respectively. OFF impedance reaches 8 M Ω .

Each of the chip's comparators has typical offsets of 3 mV (12 mV maximum) and 20 nA. A bias of 200 nA is typical. The comparators use pnp differential inputs, so that comparisons down to V^- can be performed. And slew rates of +50 V/ μ s can be achieved by connection of an external resistor.

When used for log conversion, the IC can display a variable voltage over a 50-dB range to an accuracy of ± 0.1 dB. Besides display and driver/decoder units, a complete log converter requires only one ER201M, two RC networks, a diode and a pull-up resistor.

The ER201M comes in a 16-pin DIP and operates from -55 to 125 C. In hundred quantities, units cost \$15; for 10,000 quantities, the unit price drops to \$5.75. Samples are available.

CIRCLE NO. 251

1-k TTL pROMs feature MIL temp range

Fairchild Semiconductor, 464 Ellis St., Mountain View, Calif. 94042. (415) 962-3816. \$39 (100 up); stock.

Two new versions of the company's 256 \times 4-bit TTL programmable ROMs operate over the full MIL temperature range of -55 to 125 C. The pROMs are the 93416, with open-collector outputs, and the 93426, which has three-state outputs.

CIRCLE NO. 253

64-bit ECL RAM has 6.5-ns access

Signetics, 811 E. Arques Ave., Sunnyvale, Calif. 94086. (408) 739-7700. \$19.40 (100); stock.

The 10142, a 64-bit ECL RAM, features a typical address access time of only 6.5 ns and a chip-select access time of 5.5 ns. This ECL 10-k memory is pin-for-pin and functionally equivalent to the popular 10148 RAM, which has a typical access time of 8.5 ns. Both read and write cycles for the 10142 can be accomplished in 10 ns, allowing data rates up to 100 MHz.

INQUIRE DIRECT

Divide-by-4 counter operates at 250 MHz

Plessey Semiconductors, 1674 McGaw Ave., Santa Ana, Calif. 92705. (714) 540-9979. \$7.90 to \$18 (100 up); stock.

A high-speed divide-by-4 counter series operates with a specified input frequency range of 15 to 250 MHz over the full military temperature range. The series has a low current drain of 23 mA maximum. At 25 C, typical current drain is 16 mA and typical maximum input frequency is 390 MHz. The SP600 employs external bias and it uses capacitive coupling to the signal source. Inputs can be either single, or double driven with two complementary input signals. And load resistors may be taken to any bias voltage up to 12 V more positive than V_{EE} .

CIRCLE NO. 254

Character generator interfaces to printer

Nortec Electronics Corp., 3697 Tahoe Way, Santa Clara, Calif. 95051. (408) 732-2204. \$14.50. (100-999).

A MOS/LSI ASCII-code character generator, the Model NEC-4881, interfaces directly with Bowmar's TP-3100 Thermal Printer. The 4881 decodes 7-bit ASCII code into the appropriate 5×5 matrix format for printing. The IC also has mask options available for 4×5 and 5×7 matrix outputs.

CIRCLE NO. 255

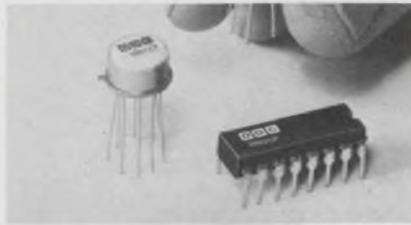
7 ICs expand MIL-temp MECL 10-k

Motorola, Semiconductor, P.O. Box 20924, Phoenix, Ariz. 85036. (602) 244-3466. \$11.25 to \$29.50 (100-999).

Seven flat-pack versions of circuits in the MECL 10,000 series are added to the MIL temperature-range lineup. The recent additions consist of a quad latch (MC-10533F), universal decade counter (MC10537F), binary 1-of-8 decoders (MC10561F/62F), dual binary 1-of-4 decoders (MC10571F/72F) and a look-ahead carry block (MC-10579F).

CIRCLE NO. 256

FET op amps come in single, dual versions



ILC Data Device Corp., 100 Tec St., Hicksville, N.Y. 11180. (516) 433-5330. \$4 to \$50; stock to 4 wks.

A line of FET-input op amps, the 1001 and 1002 series, combines an input bias of typically 2 pA with a slew rate of 6 V/ μ s. The 1001 series is pin compatible with the 741 op amp, while the 1002 series includes two 1001-type amps in a single package. Selected versions are available with a maximum input bias current of 1 pA, and offsets and drifts as low as 1 mV and 5 μ V/ $^{\circ}$ C, respectively. The 1001 series comes in a TO-99 can, while the 1002 is available in a 16-pin ceramic or plastic DIP. Inputs and outputs are protected against short circuits.

CIRCLE NO. 257

IC contains i-f part of FM receiver



Signetics, Consumer Products, 811 E. Arques Ave., Sunnyvale, Calif. 94086. (408) 739-7700. \$2.55 (100).

An advanced phase-locked loop IC, the NE563, can be used to replace many of the components presently needed in the i-f section of FM receivers. Without requiring coils, inductors or variable tuning components, the NE563 performs such functions as agc, afc, drive for a signal-strength meter and variable muting. The IC has 5-mV rms input sensitivity (with 30 dB S/N), 500-mV rms recovered audio (75-kHz deviation) and 0.05% harmonic distortion (75-kHz deviation and 1-kHz modulation).

CIRCLE NO. 258

TV sound system contained in single IC

National Semiconductor Corp., 2900 Semiconductor Dr., Santa Clara, Calif. 95051. (408) 732-5000.

The monolithic LM1808 contains the audio circuitry required by a TV receiver. The unit, a 2-W audio i-f circuit, basically combines the features of two earlier ICs—the LM3065 and the LM380. However, an improved volume-control circuit allows a single nonshielded wire to be run from the LM1808 to the front panel for volume control.

INQUIRE DIRECT

FET switch driver outputs up to 30 V

Siliconix Inc., 2201 Laurelwood Rd., Santa Clara, Calif. 95054. (408) 246-8000. \$6.80 to \$20.40 (100); stock.

A dual FET analog-switch driver comes with complementary outputs for MOS interface applications. Called the D139, the new circuit incorporates bipolar, PMOS and Schottky technology on a common substrate, and each output can drive one or more PMOS or NMOS FETs directly, or drive a junction FET switch with a few additional circuit elements. Output voltage swing is as high as 30 V and switching time is 200 ns max. Supply current is 1.5 mA typical.

CIRCLE NO. 259

400-ns 1-k RAM dissipates only 263 mW

Advanced Micro Devices Inc., 901 Thompson Pl., Sunnyvale, Calif. 94086. (408) 732-2400. \$16.95 to \$32 (100).

A high-speed, low-power 1024-bit NMOS RAM—the Am9102—offers access times down to 400 ns (suffix B). In addition, the company also has a 500-ns circuit (suffix A) and a 650-ns circuit (basic model). All versions of the 2102-type memory guarantee power dissipation at 263 mW, and they feature a dc standby mode that reduces power requirements by 75%. Fanout is guaranteed at two TTL loads.

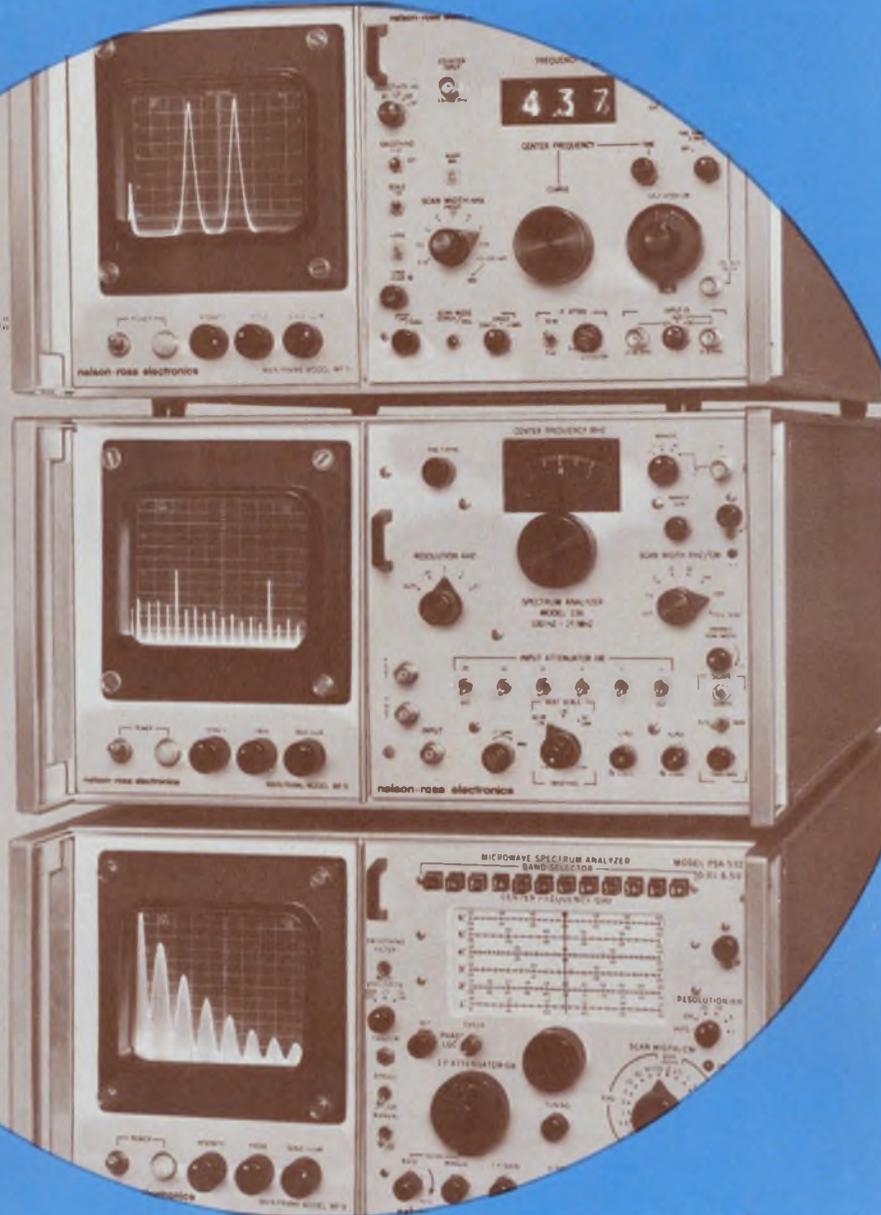
CIRCLE NO. 260

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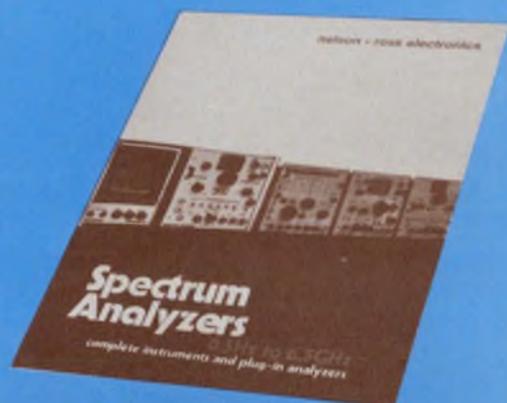
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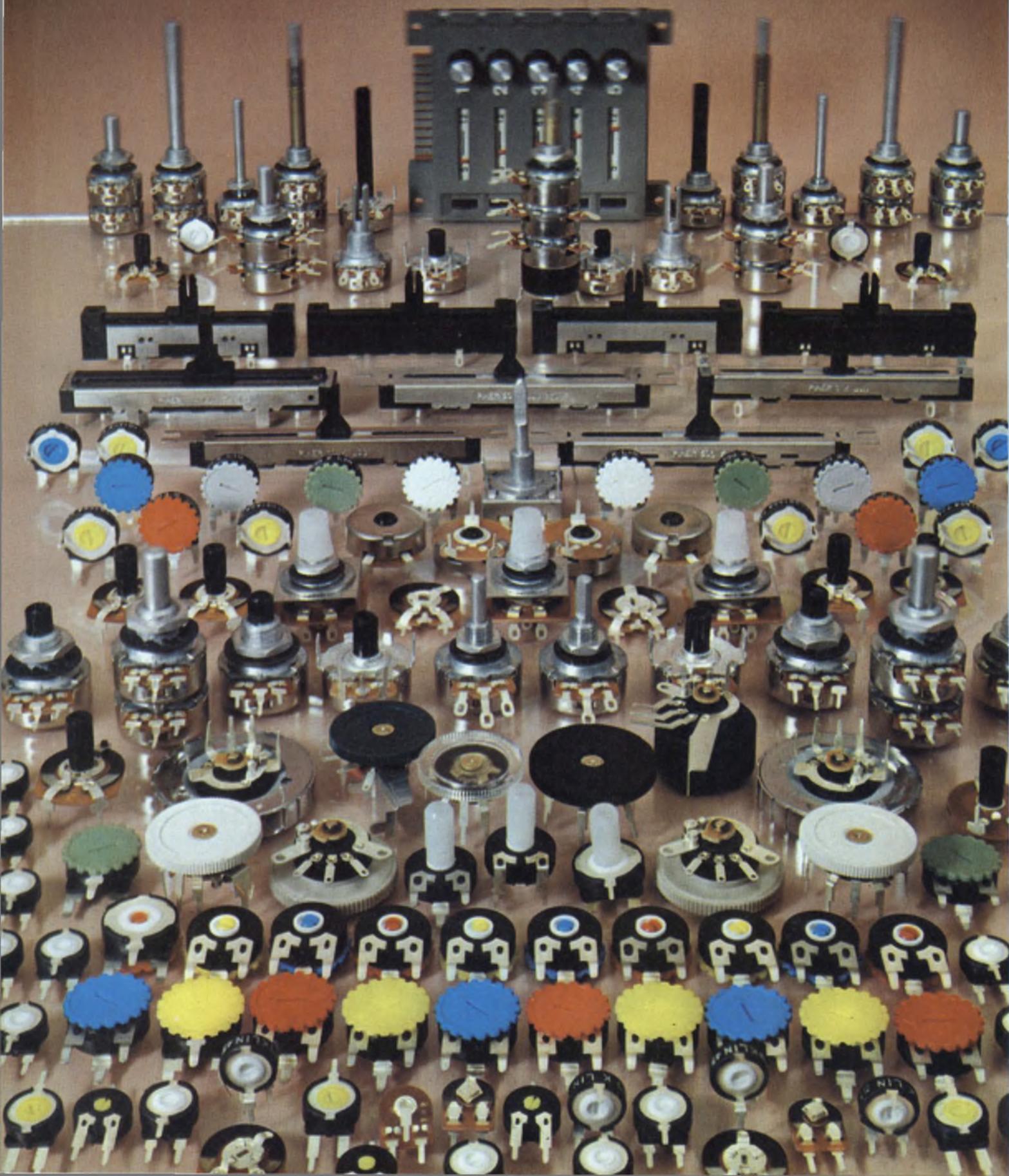
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INFORMATION RETRIEVAL NUMBER 73



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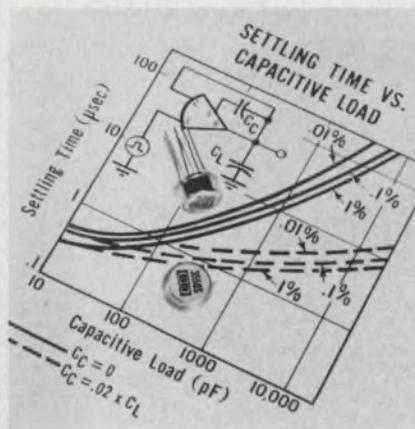
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INTEGRATED CIRCUITS

FET op amps settles in 600 ns to 0.01%



Burr-Brown, International Airport Industrial Park, Tucson, Ariz. 85706. (602) 294-1431. \$15 to \$26 (100); stock to 4 wks.

Not only does the Model 3550K FET-input op amp provide a unity-gain bandwidth of 20 MHz and a guaranteed slew rate of 100 V/ μ s. The new IC also offers a maximum settling time to 0.01% of just 600 ns. And these specs hold when the internally compensated op amp, which has a full differential input, is operated in a noninverting or inverting mode. Typical dc open-loop gain is 100 dB and input impedance is $10^{11}\Omega$. Good settling can also be obtained with large capacitive loads when a small external compensation capacitor is added. Settling with a 10,000-pF load then reduces by over an order of magnitude.

CIRCLE NO. 517

Dual, quad op amps use single supply

Signetics, 811 E. Arques Ave., Sunnyvale, Calif. 94086. (408) 739-7700. \$1.05 to \$2.50 (100); stock.

Both the NE532 dual op amp and the LM324 quad op amp can operate from a single source of 3 to 30 V. Operation from dual supplies of 1.5 and 15 V is also possible. Internally compensated, differential voltage amplifiers are used in both ICs. Each op amp has a unity-gain bandwidth of 1 MHz and a bias current of 45 nA, both of which are temperature compensated. Input offset voltage is 2 mV.

INQUIRE DIRECT

4-k-bit ROM accesses in 330 ns

Siemens, Z1/Presseabteilung Technik, D-8520 Erlangen, Postfach 3240, Germany.

The latest 4096-bit RAM has an access of 330 ns, a read-refresh cycle of 600 ns and a write cycle of 800 ns. Packaged in a 22-pin DIP, the Model S 142 memory dissipates 400 mW in operation and 10 mW when idle. Supplies required for the NMOS RAM are 12, 5 and -5 V. Refresh for the dynamic memory takes 64 cycles, and its operating temperature range extends from 0 to 70 C.

CIRCLE NO. 518

ICs performs sample and hold function

Intersil, Inc., 10900 N. Tantau Ave., Cupertino, Calif. 95014. (408) 257-5450.

Two sample-and-hold circuits are being offered. The IH5110 is optimized for signals up to 10 V ac pk-pk or ± 10 V dc, and the IH5111 is optimized for signals up to 20 V ac pk-pk or ± 10 V dc. Drift rate for both devices—with an external 0.01- μ F capacitor—is 10 mV/sec maximum, and less than 2 mV/sec typical. Maximum charge injection is 5 mV pk-pk for the IH5110 and 10 mV pk-pk for the IH5111. Maximum acquisition time for a 10-V step input is 10- μ s for the IH5110 and 15 μ s for the IH5111. Aperture time for both units is 500 ns.

CIRCLE NO. 519

MOS terminal IC aids data systems

Motorola Semiconductor, P.O. Box 20924, Phoenix, Ariz. 85036. (602) 244-3466. \$10.75 (100-999); stock.

The MC2260L, an MOS asynchronous/synchronous terminal transmitter, provides a serial communications link in data systems. Character lengths are selectable in 5, 6 or 7 bits. Also, one or two stop bits are externally selectable in asynchronous mode. Data rates are from dc to 10,000 bits per second (BPS) in a divide-by-64 mode, 40,000 BPS in a divide-by-16 mode and 200,000 BPS in the divide-by-1 mode or at the clock frequency.

CIRCLE NO. 520

MORE POWER MORE BANDWIDTH

Our Model 503L is an ultra-wideband RF power amplifier whose wide range of frequency coverage and power output provides the user with the ultimate in flexibility and versatility in a laboratory instrument. Easily mated with any signal generator, this completely solid state unit amplifies AM, FM, SSB, TV, pulse and other complex modulations with minimum distortion.

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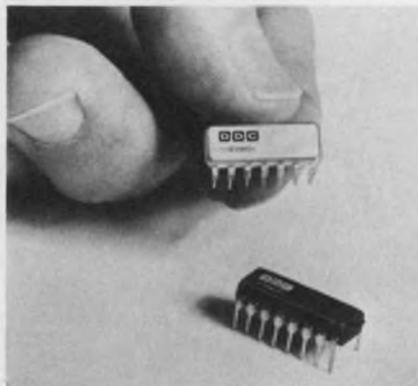
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INFORMATION RETRIEVAL NUMBER 75

INTEGRATED CIRCUITS

FET analog switches handle 20-V signals



ILC Data Device Corp., 100 Tec St., Hicksville, N.Y. 11801. (516) 433-5330. Less than \$1 per switch; stock to 4 wks.

A complete line of FET analog switches (the 1101 through 1108 series) are available in both industrial and military-temperature ranges. Each package contains up to four channels of analog switches and eliminates the need for an external driver. The 1101 through 1104 devices are driven directly from low-level TTL logic (5 V), and can switch signals in the -10 to +10-V range. The 1105 through 1108 series is driven from TTL open-collector logic (15 V) and can switch 0 to 20-V signals. Each channel of the 1101 through 1108 simulates a SPDT switch, while that of the 1105 through 1108 simulates a SPST switch. The ON resistance for each channel is less than 150 Ω .

CIRCLE NO. 261

Static registers guarantee 1.5-MHz rate

Advanced Micro Devices Inc., 901 Thompson Pl., Sunnyvale, Calif. 94086. (408) 732-2400. \$4.80 to \$9.20 (100).

Three static shift-register circuits have speeds guaranteed to 1.5 MHz. The trio of p-channel, silicon-gate MOS units include these: the Am40/5055, a quad 128-bit circuit; the Am40/5056, a dual 256-bit circuit; and the Am40/5057, with 512 bits. These devices are pin compatible with like-numbered circuits originally offered by National Semiconductor.

CIRCLE NO. 262

ECL counters operate above 100 MHz

Signetics, 811 E. Arques Ave., Sunnyvale, Calif. 94086. (408) 739-7700. \$13.55 (100).

Two ECL counter circuits, the 10136 and the 10137, offer operating speeds in excess of 100 MHz. The 10136 is a 16-state hexadecimal synchronous counter, while the 10137 is a 10-state, or decimal counter. Both devices have four operating modes—count up, count down, preset and stop—controlled by two function-select inputs. Other features include a carry-in and carry-out for extending the count range, a clock input for synchronous operation, and four data inputs for presetting the counter and variable-modulus count operations.

INQUIRE DIRECT

Isoplanar RAMs switch to plastic

Fairchild Semiconductor, 464 Ellis St., Mountain View, Calif. 94042. (415) 962-3816. \$12.40 to \$22 (100); stock.

Plastic-packaged versions are offered for three Isoplanar RAMs, including the company's 93415 TTL 1024-bit RAM. The two other RAMs are the 93411 TTL 256 \times 1-bit memory (with open-collector output) and the 93421 TTL 256 \times 1-bit memory (with three-state output).

CIRCLE NO. 263

CMOS decoder/driver aims for LCD watch

Nortec Electronics Corp., 3697 Tahoe Way, Santa Clara, Calif. 95051. (408) 732-2204. \$17.50 (100-999).

The NEC-5015 CMOS decoder drives a 3-1/2 digit, seven-segment field-effect LCD directly without external components. The new decoder/driver handles a display of hours and minutes from 23 outputs and a common line with a 32-Hz ac signal. Seconds are displayed with a flashing colon, and two time-setting inputs are provided for hours and minutes. The new 5015 can interface with an oscillator/countdown chip (NEC-5016) to form a complete watch circuit.

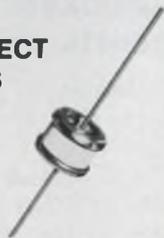
CIRCLE NO. 264

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First floppy disc peripheral made for microcomputers



iCOM, 6741 Variel Ave., Canoga Park, Calif. 91303. (213) 348-1291. See text; 45 days.

In response to the need for peripherals for microcomputers, iCOM has developed the first microperipheral floppy-disc drive with integral controller.

The disc drive, the FD360, can operate under directions received from either the Intel (Santa Clara, Calif.) or the National Semiconductor (Santa Clara, Calif.) microprocessor systems.

The FD360 is available in either a single or dual-disc version, and can be expanded so that four drives can be manipulated by one controller. Two drives fit into the main chassis, and up to two more drives can be put into an expansion chassis.

Hardware interfaces are available for the Intel Intellec-8 and Intellec-8/MOD-80 and National's IMP-16P, 16L and 8P.

The manufacturer uses Pertec's IBM-compatible floppy-disc drives in the system. The floppy discs thus have format compatibility with the IBM 3741, 3742 and 3540 systems.

Features of the FD360 include

the following: motor shutdown to reduce disc hub wear; built-in hardware track seek and seek verification; automatic head unload and load to minimize head and media wear; individual write-protect switch for each drive; operation with either programmed I/O or DMA interfaces; automatic single-command write of deleted data-address mark, and input and output buffering to enable asynchronous programmed I/O.

Input and output buffering (128-byte buffers) on the FD360 allows integration of the microperipheral with other asynchronous data handling devices such as keyboards, communications equipment, paper tape and data-acquisition equipment.

Built-in CRC (cyclic redundancy code) generation and verification hardware works with the data buffers to make write verification unusually easy on the FD360. When data are written onto the disc from the write buffer, the data are recycled into that buffer. The written data are then verified when a "read for CRC verification" command is issued. This function reads, checks and discards the data

being read, thus not destroying the existing contents of the read buffer. If a data error is detected, only a new write command need be issued, since the data are still present in the write buffer.

Interfacing to the FD360 is done through eight input lines and 16 output lines.

The drive and controller are contained in a 10 × 17 × 17-in. desk-top cabinet, and the system uses 115 V ac, 47 to 440 Hz lines and consumes 150 W.

In a single drive configuration, the selling price is \$2350 (unit quantity). With two drives, the price rises to \$3000. Special interfaces cost a few hundred dollars more.

In addition to the microcomputer interfaces, iCOM also offers hardware to interface the system to the Hewlett-Packard 2100 and Data General Nova families of mini-computers.

CIRCLE NO. 252

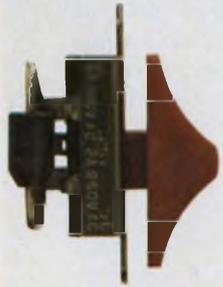
Arithmetic logic unit boosts mini throughput

Interdata, 2 Crescent Pl., Oceanport, N.J. 07757. (201) 229-4040. \$4900.

An outboard arithmetic logic unit dubbed HSA LU improves performance of the Model 7/16 mini. The unit provides hardware multiply/divide, floating point arithmetic and list processing. The latter assists real-time schedulers by indicating the number of slots in a list and the number used as the lists are processed. The unit also detects privileged instructions—a feature found mainly in large-scale computer operating systems. According to the manufacturer the unit improves standard execution times by 30% or even more.

CIRCLE NO. 265

Our switches are very sensitive to your needs



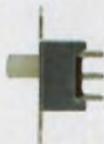
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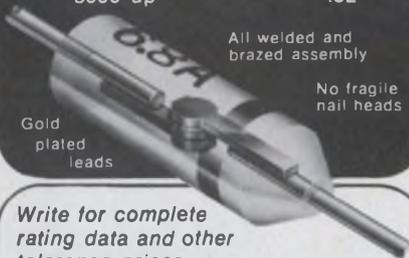
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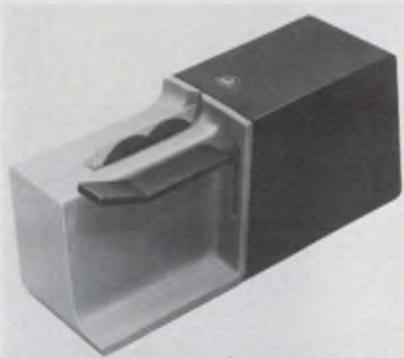
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DATA PROCESSING

**Credit-card reader
is hand operated**

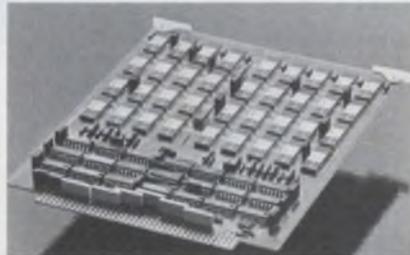


American Magnetics Corp., 2424 Carson St., Torrance, Calif. 90501. (213) 775-8651. Under \$50 (quan).

The Model 10 reader accepts information from magnetic-stripe cards and takes about two seconds to read a card. The unit is hand operated. A spring and a dash pot provide a constant read velocity so stray magnetic fields that accompany the use of an electric motor are eliminated. Double-axis gimbals mount the magnetic read head and supply the necessary compliance to read warped cards.

CIRCLE NO. 266

**NMOS memories offer
275-ns access times**



Electronic Memories & Magnetics Corp., 12621 Chadron Ave., Hawthorne, Calif. 90250. (213) 644-9881. \$495 to \$617; 40 days.

Basic systems in the Microram NMOS memory series are organized as 4096, 9-bit words. Board size is 7 x 8.875 in. for the N models and 7 x 10.75-in. for the NV models. The NV models require only 5 V; the N models use 15 V, -5 V and 5 V. The 1200 N and NV units provide 350 ns access and 900 ns cycle times. The 1210 N and NV units have 275 ns and 400 ns access and cycle times. All memories feature TTL compatible operation.

CIRCLE NO. 267

**Modem for 4800 bit/s
is Bell 208B compatible**



Rixon, 2120 Industrial Parkway, Silver Spring, Md. 20902. (301) 622-2121.

The T208B modem provides 4800 bit/s data rate and is end-to-end compatible with the AT&T 208B unit. The unit operates over two-wire direct-dial facilities in the half-duplex mode and is equipped with automatic equalization. Features include automatic answer, automatic echo suppressor disable and alternate voice/data.

CIRCLE NO. 268

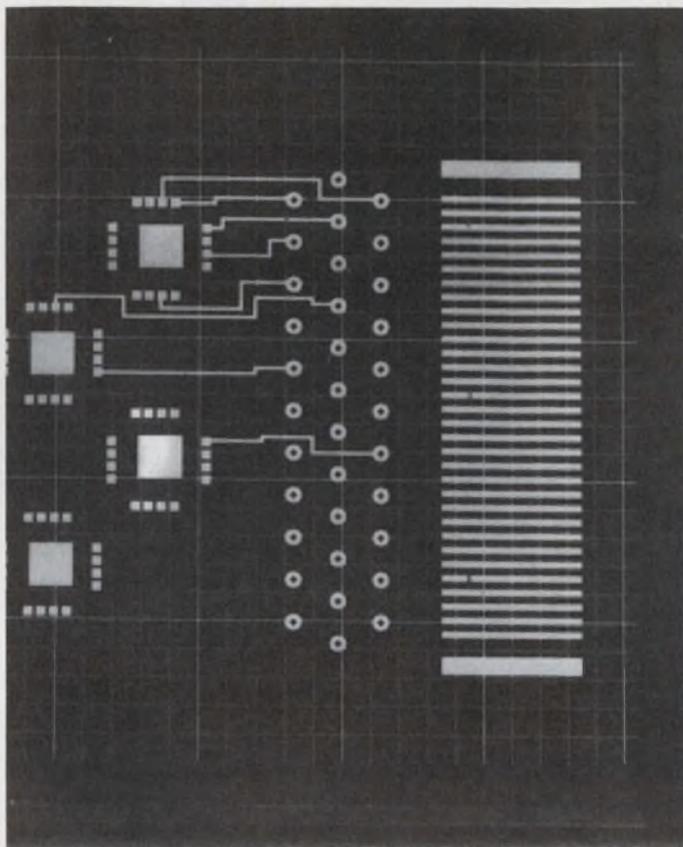
**Smart controller is
fast and programmable**

Xerox Corp., 701 S. Aviation Blvd., El Segundo, Calif. 90245. (213) 679-4511. From \$4500.

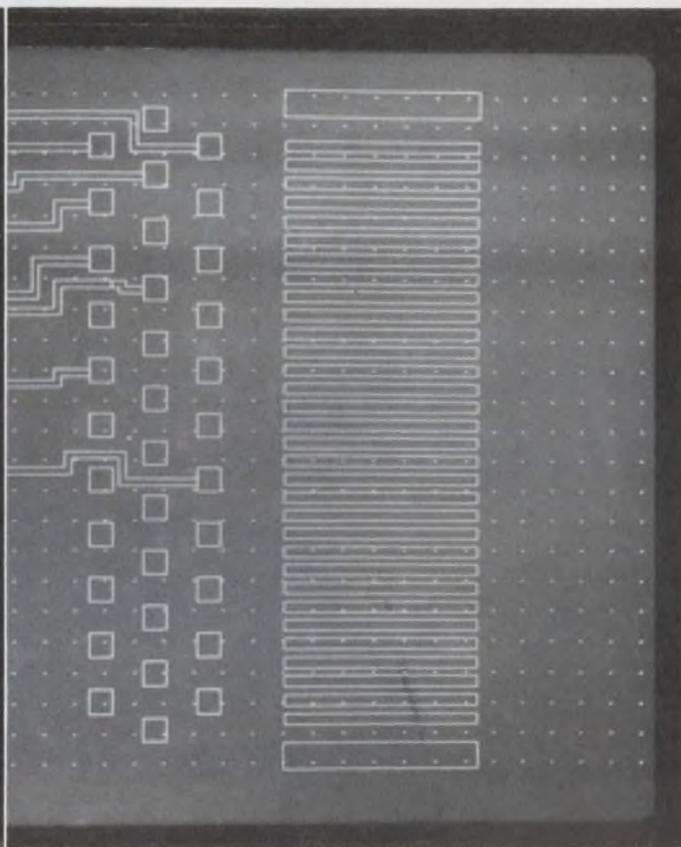
A microprogrammable controller, dubbed the system control unit (SCU), has a throughput rate of 2.86 M, 16-bit word/s. The device is designed to provide high data throughput and fast response to interrupts. Response to interrupts is 350 ns and the multiplexed interrupt response is as fast as 1.05 μ s. SCU applications include use as an I/O controller, preprocessor for a host computer, device controller for peripherals, communications system controller, special algorithmic processor, computer emulation system, remote intelligent processor, video display controller, and as a stand-alone processor. A three-bus 16-bit parallel architecture allows up to three concurrent data transfers within a single microinstruction. Data may be intermixed combinations of bytes and 16-bit words. Other features of the SCU include 256 to 4096 words of 32-bit control memory, 4000 to 65,000 words of 16-bit data memory, eight general-purpose registers, and up to 128 I/O addresses. The unit is available with a wide variety of standard peripheral interfaces as well as a general I/O structure.

CIRCLE NO. 269

Digivue® - a better way to look at it.



Unretouched time exposure of data displayed on a Digivue unit. Note high contrast picture, precise graphics.



Unretouched time exposure of same data displayed on a CRT. Note lack of contrast in CRT image.

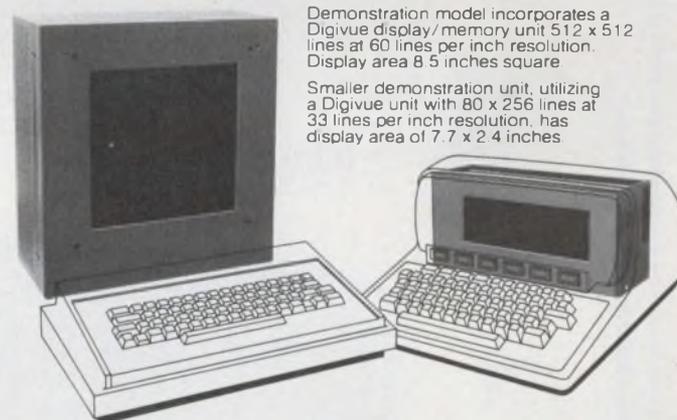
When your customers spend a lot of time looking at data displays and computer terminals, the advantages of Digivue plasma display units over CRT's become clear.

Very simply, Digivue display/memory units are better to look at. Digivue units present a flicker-free high contrast display for more precise readings and reduced chance of eye fatigue.

And Digivue units make it easier to use what's on the screen. Inherent memory is a feature of every unit. Digivue images appear on a thin, flat panel—which allows you to design compact components. The 512-60 models offer hard copy and rear projection capabilities, too. All of which gives you a product with more features to sell.

Sure, Digivue display/memory units currently cost more than CRT's. But they're worth a lot more to your customer, because they do a lot more.

For an informative booklet about Digivue display/memory units call (419) 242-6543, Ext. 66-415, or write Electro/Optical Display Business Operations, Owens-Illinois, Inc., P.O. Box 1035, Toledo, Ohio 43666.



Demonstration model incorporates a Digivue display/memory unit 512 x 512 lines at 60 lines per inch resolution. Display area 8.5 inches square

Smaller demonstration unit, utilizing a Digivue unit with 80 x 256 lines at 33 lines per inch resolution, has display area of 7.7 x 2.4 inches

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Toledo, Ohio

Now New from T.R.I. a engineering and

All four of these new additions to our product line will cause quite a stir. They're totally new and so is a lot of the technology behind their manufacture. Technology developed specifically to meet user needs and innovate user measuring techniques.

There's also something else behind their manufacture in addition to new technology. T.R.I.'s traditional concept of measuring instrument efficiency, reliability, easier use, and increased cost economy.

Find out about the total advantages of all four direct. Phone (408) 733-9080, 9081. Or write us. We'll send complete details to supplement the specs we're introducing here.

Price **\$279.00**



Model: 6355 Mini-multimeter

- Five functions. Automatic range selection.
- Drop-proof, portable (4-3/8W, 2-7/8H, 6D in.), Lightweight (1.8 lb)
- High performance—DCV measurements at 10 μ V resolution. ACV measurements up to 1000Vrms.
- Operates on DC battery and AC power supply.

10 μ V, Fully automatic DMM. \$279

Announcing the fully automatic, 10 μ V resolution Model 6355 portable DMM standard. And at only \$279. Not only that, but also 5 functions and automatic range selection. Measures AC voltages up to 1000Vrms, can even be operated from a battery. The same performance as a laboratory use DMM. Can be easily hand carried. Weights only 1.8 lb. Moreover, shock resistant design protects it even if it is accidentally dropped. Five auto functions range switching, units display, polarity, overrange indication, and overload protection.

Display: 1999 **DC Voltage:** 20 mV—1000 V (resolution 10 μ V) **AC Voltage:** 200 mVrms—1000Vrms (resolution 100 μ V) **Resistance:** 200 Ω ~ 20M Ω (resolution 100m Ω) **DC Current:** 20 mA ~ 1A (resolution 10 μ A) **AC Current:** 200 mA ~ 1A (resolution 100 μ A) **Size:** Approx 4-3/8 (W) x 2-7/8 (H) x 6 (D) in. **Weight:** Approx 1.8 lb **Price:** \$279

Repetition rate **500MHz**



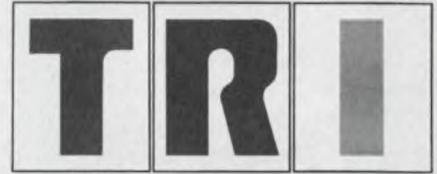
Model: 4220 Pulse Generator

Maximum repetition rate up to 1000 MHz make it ideal for PCM communication.

Model 4220 is a high-performance pulse generator with a repetition rate of 500 MHz. It can also generate word patterns. Among the main features of this high performance instrument is generation of a multitude of waveforms through combinations of different modules. Eight modules enable pulse generator tailoring to a specific requirement. Each unit also has a remote control function and such complementing features as delay of 1 or more periods, 50 Ω output impedance which enables generation of highly accurate pulses, variable rise and fall time, and two outputs.

Repetition Frequency: 100 kHz—500 MHz **Modules:** Delay, double pulse gate, word generator, word/pseudo random noise generator, NRZ output amp, RZ output amp, RZ output amp (rise time/fall time variable), bipolar output amp.

Product Line to Innovate production ideas.



T.R.I. CORPORATION
505 West Olive Avenue Sunnyvale, CA 94086
(408) 733-9080

Programmable



Model: 6141
Programmable DC V/I Generator

- Economy plus programmable functions suitable for a host of OEM applications.
- Easy flip-switch operation.
- Extremely stable. Also has pulse-width modulation system.
- Variable in $1\mu\text{V}$ steps. All-range continuous-function performance.

Suitable for OEM and production line use.

Model 6141 features a remote control function for OEM use and wide usage on production lines. These unique features have been made possible through the adoption of a new pulse-width generator system. Easy usage has been realized through the adoption of a flip switch to automatically set the output level with one touch instead of the conventional rotary switch. Since automatic carry is performed when this switch is depressed, automatic sweep is also possible.

DC Voltage: $0\sim\pm 11.999\text{V}$ ($1\mu\text{V}$ steps) **Stability:** $\pm 0.015\%$ of setting $\pm 3\mu\text{V}$ (10 mV range) **DC Current:** $0\sim\pm 119.99\text{mA}$ ($0.1\mu\text{A}$ steps) **Stability:** $\pm 0.02\%$ of setting $\pm 0.2\mu\text{A}$ ($1\mu\text{A}$ range) **Dimensions:** Approx 11-3/4 (W) x 3-5/16 (H) x 10 (D) in. **Weight:** Approx 11.023lb **Price:** \$890

Resolution 0.00001Hz



Model: 5112/13
Milli-Hertz Counter

High resolution counter which permits direct reading of rotational speed and flow rate.

Model 5112 and 5113 are revolutionary products which permit direct-reading measurement of rotational speed, flow rate, and other comparatively low frequencies (1Hz~1MHz) at high accuracy in these respective units. Moreover, measurement accuracy has been increased through the averaging function. The measured result can be directly displayed and a multiplier function is employed to directly read physical quantity.

Frequency Range: 1Hz~1MHz **Resolution:** 0.1mHz (Model 5112), 0.01mHz (Model 5113) **Display:** 199999 (5112), 1999999 (5113) **Size:** 8-7/16 (W) x 4-2/4 (H) x 12-15/16 (D) in. **Weight:** Approx 11.02lb **Price:** \$1300 (5112), \$1400 (5113)

TO-5's WE CAN COOL 'em



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1134-U Press on
Typical R_{θ} , 59° C/W



5.1¢*

2226-U Press-on
Clears adjacent compo-
nents
Typical R_{θ} , 59° C/W



8.2¢*

2230-B5, 3/4" Press-on
gear type, screw mount
Typical R_{θ} , 51° C/W



8.5¢*

2257-R 1/2" Press-on
gear type
Typical R_{θ} , 68° C/W

Popular, easy-to-use economy coolers for PC board components. Thermalloy offers the most complete line of heat sinks in the industry: over 100 designs for cooling all types of packages including DIP's, plastics, TO-3's, TO-66's, and power transistors.

*Prices for 5000 quantity
 R_{θ} values for natural convection @ 1 W

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DATA PROCESSING

On-line storage for minis stores 262-k bytes



Kennedy Data Systems, 31829 La Tienda Dr., Westlake Village, Calif. 91361. (213) 889-4455. Under \$3300.

The Model KDS 28C floppy-disc memory includes two drives, controller cable and hardware and is designed for use with Computer Automation's Alpha 15, LSI-1, LSI-2 and Naked Mini/LSI processors. The system affords 0.5-s random access to 262 kbytes (131 k for each disc) with a transfer rate of 3.7 kbyte/s. Track-to-track access time is 6 ms. The price includes software drivers. System size is 5.25 x 19 x 22 in.

CIRCLE NO. 270

Low-cost CRT terminal displays 960 characters



Digital Equipment Corp., Components Group, One Iron Way, Marlborough, Mass. 01752. (617) 481-7400. \$950 (100 quan).

The VT50 terminal displays 12 lines of 80 dot-matrix characters and can operate at rates up to 9600 bit/s. A cursor underscores the selected character and can be moved a line-at-a-time or a character-at-a-time. Transmission code is ASCII with a 20-mA current loop. An EIA interface is optional. A separate electrolytic copier is available to record the screen contents.

CIRCLE NO. 271

Core memories available with 256-k words

Standard Memories, 2801 E. Oakland Park Blvd., Fort Lauderdale, Fla. 33306. (305) 566-7611. From \$49,000; 60 days.

The ECOM H-Series core memories use a coincident current three-wire 3D technique that provides full cycle times of 750 ns and allows access to data in 325 ns. Maximum capacity is 262-k, 20-bit words. Temperature compensation is provided from zero to 50 C and data save circuits are included. Other models are available with word sizes of 40-bits and capacities to 262-k words.

CIRCLE NO. 272

Laser system sends data at 9600 bit/sec



International Laser System, Op Com Div., 2111 W. Central Blvd., Orlando, Fla. 32805. (305) 843-4731. From \$7200; 4 mos.

An injection laser is part of an optical system that provides 9600 bit/s asynchronous transmission for line-of-sight distances up to 3000 ft. Two of the four models in the DL-100 series provide a voice channel that is multiplexed with data on a single carrier. The terminals are compatible with current-loop or RS-232-C interfaces. Warehouses, manufacturing facilities and refineries are places where the system is said to offer advantages over wire lines or rf links.

CIRCLE NO. 273

Calculator simplifies technical computations

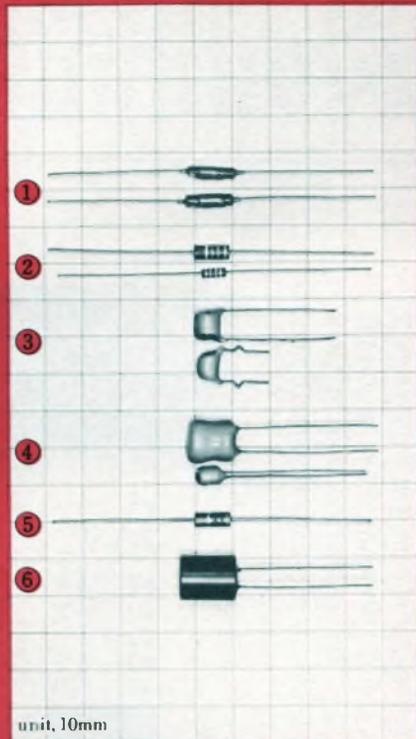
Canon U.S.A., 10 Nevada Dr., Lake Success, N.Y. 11040. (516) 488-6701. \$169.95.

The Palmtronic F-5 calculator performs a variety of scientific calculations that include trigonometric, exponential logarithmic and square root. The calculator also performs the four basic functions and various mixed calculations. Other features include floating point, 8-digit display and memory to retain intermediate calculations.

CIRCLE NO. 274

GUIDEPOST FOR TDK FIXED INDUCTORS

TDK fixed inductors you may select from the table are all production units supported by comprehensive quality assurance program in various phases including own ferrite, coil winding, expertise core design and manufacturing technique.

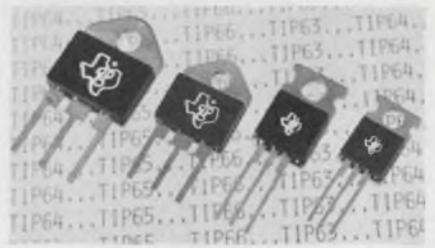


TDK
 TDK CORPORATION OF AMERICA
 931 South Douglas St., El Segundo, Ca 90245 Phone: (213) 644-8625
 CHICAGO BRANCH
 2906 West Peterson Av., Chicago, Ill 60659 Phone: (312) 973-1222
 NEW YORK BRANCH
 755 Eastgate Blvd., Garden City, N.Y. 11530 Phone: (516) 746-0880
 EL PASO BRANCH
 Room 6, 8201 Maritana Place, El Paso, TX 79925
TDK ELECTRONICS CO., LTD.
 14-6, 2-chome, Uchikanda, Chiyoda-ku, Tokyo 101, Japan

| | | Type & photo No. | Features | Size (mm) $\phi \times l$. | Inductance range | Nr. of L-values available | Q (min.) range | Catalog No. | | | | |
|--|-------------|------------------|--|-----------------------------|--------------------------|---|----------------|-------------|------------------|----|----|------------|
| Magnetically Non-shielded Construction | Axial Lead | Type SP ① | Small size, low dc resistance Inductor's figure of merit sustained well into high frequency range because of low distributed capacitance | 4 x 6 | 0.82 μ H-1mH | 55 | 30-85 | BCE2X-013A | | | | |
| | | | | 4 x 8 | 0.15 μ H-27 μ H | 25 | 45-65 | | | | | |
| | | | | 4 x 10 | 0.15 μ H-56 μ H | 31 | 45-75 | | | | | |
| | | | | 4 x 12 | 0.15 μ H-91 μ H | 36 | 45-80 | | | | | |
| | Radial Lead | Type TP ② | Packaging in transfer molding High reliance design to meet MIL specifications | 2 x 6 | 0.15 μ H-100 μ H | 35 | 25-55 | BCE39-003B | | | | |
| | | | | 4 x 10 | 0.15 μ H-1mH | 62 | 33-75 | | | | | |
| Magnetically Shielded Construction | Axial Lead | Type TPF ⑤ | Magnetically shielded by TDK ferrite High inductive loading factor Packaging in transfer molding Performance characteristics to meet MIL specifications | 4 x 10 | 0.1 μ H-100mH | 73 | 18-67 | BCE39-011B | | | | |
| | | | | Radial Lead | Type SL ④ | Elements having large inductance value Epoxy hermetic sealed | 8 x 10 | | 0.22 μ H-1mH | 89 | 30 | BCE39-012B |
| | | | | | | | 10 x 10 | | 1.1mH-2.2mH | 8 | 30 | |
| Magnetically Shielded Construction | Radial Lead | Type FS ⑥ | Type SL elements in ferrite shielded configuration Best used in circuits critical of electromagnetic induction hazards | 6 x 9 | 1 μ H-100 μ H | 24 | 45-80 | BCE39-016A | | | | |
| | | | | 8 x 12 | 100 μ H-8.2mH | 28 | 60-150 | | | | | |
| | | | | 13 x 16 | 10mH-100mH | 16 | 60-140 | | | | | |
| | | | | 8 x 10 | 12 μ H-1mH | 41 | 55-95 | BCE39-011B | | | | |
| | | | | 10 x 12 | 1.1mH-5.6mH | 18 | 98-112 | | | | | |
| | | | | 12 x 15 | 6.2mH-100mH | 29 | 50-105 | | | | | |

Detach here and keep it in your file

Power transistors handle 300 to 1400 V, V_{ce0}



Texas Instruments Inc., Inquiry Answering Service, P.O. Box 5012, M/S 308, Dallas, Tex. 75222. (214) 238-3741. From \$0.55 to \$1.53 (100-up); 8 wk.

The two series of TIP63, TIP64 and TIP65, TIP66 npn power transistors are available in TO-66 and TO-3 plastic packages, respectively. The TIP63 and TIP64 units are high-voltage, medium-power transistors. These devices feature V_{ce0} breakdown voltages ranging from 300 to 350 V and continuous power dissipation of 20 W at 25-C case temperatures. Other features include a free-air continuous power dissipation of 2 W at 25 C and 0.5-A continuous collector current. The TIP65 and TIP66 units are horizontal TV deflection transistors designed for line-operated CRT deflection circuits. They have 1200 and 1400 V collector-emitter off-state voltage ratings and 1.5-A collector currents. Their switching fall times are 0.7 μ s at a current of 1 A typically.

CIRCLE NO. 275

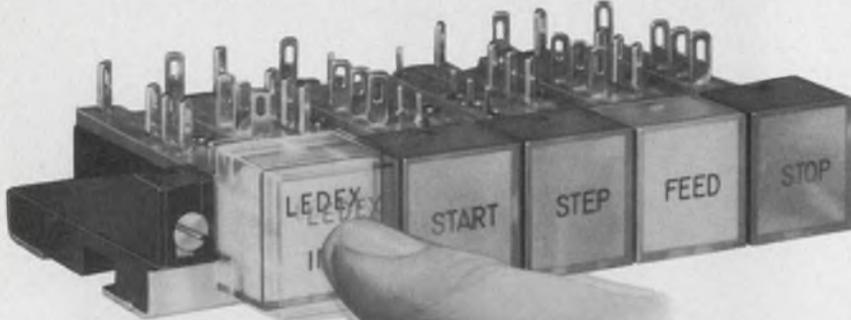
Optocouplers guarantee 2000 V of isolation

National Semiconductor, 2900 Semiconductor Dr., Santa Clara, Calif. 95051. (408) 732-5000. 1000 up lots: NCT 200 (\$1.65); 260 (\$1.20); stock.

The NCT 200 and NCT 260 optocouplers are pin-for-pin compatible with competitive devices (MCT2; MCT26; ISO-LIT 16, H11A2; 4N26; 4N27 and FCD-820). They offer a higher guaranteed isolation voltage of 2000 V minimum. The NCT 200 has a typical 80% current transfer ratio and 0.5 pF isolation capacitance. The NCT 260 has a 6% current transfer ratio, and typical 0.5-pF isolation capacitance.

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SWITCHING...



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Single illuminated push button switches, or mounted arrays on a single frame, Ledex can supply your switching needs through the dramatically different Series 1400 switches. You can save dollars and assembly time and have all the variations you need... Momentary... Push/Push... Reciprocal Release... Master Release... Interlock... you name it.

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Toll free number for name of your nearest representative: 800-645-9200



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Dayton, Ohio 45401
(513) 224-9891



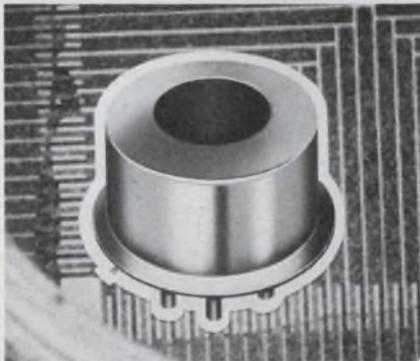
High voltage rectifiers handle up to 10 kV PIV

Sarkes Tarzian, 415 N. College Ave., Bloomington, Ind. 47401. (812) 332-1435. Under \$1 single unit.

The F60-F-100 series of high voltage rectifiers have PIV ratings of up to 10,000 V. Each model is housed in 0.335 in. long by 0.195 in. diameter case. The diodes have a maximum one cycle surge current of 30 A and are available in five values from 6000 to 10,000 V PIV. Maximum current at 60-C oil temperature is 220 mA dc. The oil bath normally used to prevent arcing with this type of device is adequate for all cooling requirements.

CIRCLE NO. 276

Thermopile detector has 120-junction active area



Sensors, Inc., 3908 Varsity Dr., Ann Arbor, Mich. 48104. (313) 973-1400. \$126 (unit qty).

The L66 thermopile detector has a 6-mm thin-film active area with 120 junctions. It has a responsivity of 1 to 2 V/W, making the detector an ideal candidate for laser measurement. The standard L66 is packaged in a TO-8 case, and is back-filled with argon at one atmosphere to yield D^* (at 500 K, dc) of 0.5 to 1×10^8 cmHz^{1/2} W. Although standard window material is KBr, other materials can be supplied, including: CaF₂, BaF₂, Ge, KRS-5, Irtran-2, and SiO₂. The L66 can also be supplied without the window material. Additional L66 specifications include: Noise equivalent power = 10^{-10} W/Hz^{1/2}; response time (1/e) = 60 ms; resistance = 30 to 40 k Ω ; operating temperature range = 60 to 100 C; and damage threshold = 200 mW/cm².

CIRCLE NO. 277

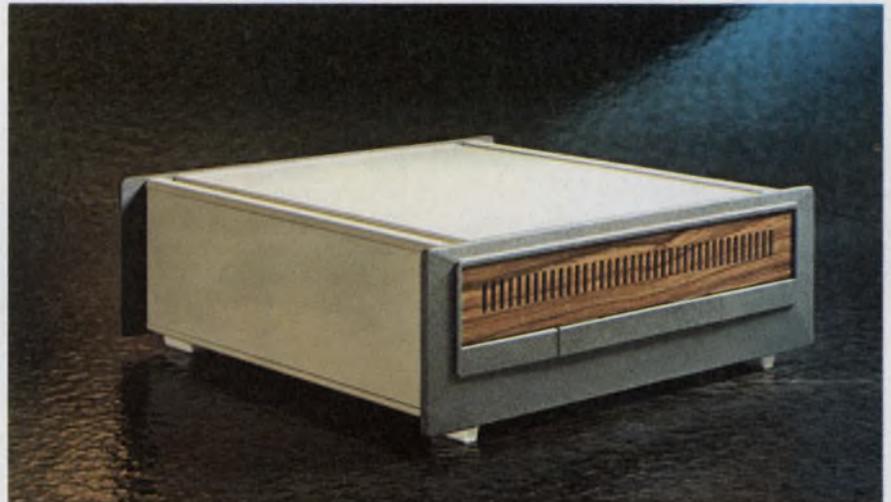
Varactor tuning diodes have Qs up to 5000

Varian, 611 Hansen Way, Palo Alto, Calif. 94303. (415) 493-4000. From \$75; stock to 2 wk.

A line of improved performance high-Q varactor tuning diodes offers guaranteed minimum values for Q varying from 2400 to 5000, and capacitance ratios up to 6:2. These GaAs device improvements

represent a 10 to 25% increase in capability. Typical applications include wide tuning range vhf and uhf oscillators, broad tuning range filters, high stability oscillators, and tuning of microwave transistor oscillators. Total capacitance values may be specified from 0.5 to 5 pF with $\pm 10\%$ tolerance. Additional values and tightened tolerances are available on special order.

CIRCLE NO. 278



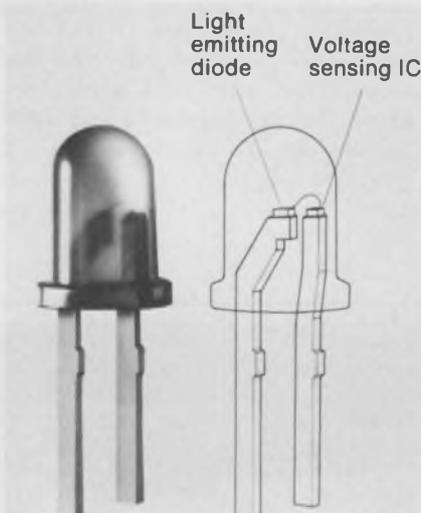
The Optima Experience

The Optima experience—it's colorful, stylish and now a great new way to package your electronics. We call it OptimaPAK. There is nothing like this case on the market today. Molded sculptured sides with handles neatly designed in... easily removed top and bottom panels and that's not all. The front and rear panels are part of a chassis assembly that slides right out of the front of the case or you may just want to mount it in a rack. We are eager to tell you all about OptimaPAK and share the Optima experience right now

OPTIMAPAK

Write Optima Enclosures, a division of Scientific-Atlanta, Inc.
2166 Mountain Industrial Blvd., Tucker, Georgia 30084 or
call (404) 939-6340.

LED includes voltage sensing circuit



Litronix, 19000 Homestead Rd., Cupertino, Calif. 95014. (408) 257-7910. \$0.60 (1000-up).

The RLC-400 is a tiny LED lamp that acts as a self-contained battery status indicator for portable equipment. The lamp gives users an early warning of imminent battery failure. It is brightly lit at 3 V and completely dark when battery voltage has dropped to 2 V. The LED RLC-400 combines a voltage-sensing integrated circuit and a GaAsP LED in a conventional T-1 size lamp package.

CIRCLE NO. 393

Power SCR handles 1.6 kA at up to 1.7 kV



National Electronics, Geneva, Ill. 60134. (312) 232-4300.

The NL-F701 is a SCR rated for 1600 A rms. It uses a disc design and is available in voltage ratings of up to 1700 V. Peak one-cycle surge forward current is 18,000 A. Turn-on current rate-of-rise, di/dt , is 100 A/ μ s. Peak gate power dissipation, $P_{g(m)}$, is 400 W for a 10 μ s gate pulse width.

CIRCLE NO. 394

Yellow and green LEDs come with axial leads

Dialight, 203 Harrison Pl., Brooklyn, N.Y. 11237. (212) 371-8800. \$0.55 (1000-up); stock.

Two miniature LED lamps with axial leads are available in green (521-9206) or yellow (521-9207) and are otherwise identical to the most popular red LEDs. Maximum ratings for these lamps are: power dissipation of 112 mW, derate linearly from 25 C at 1.49 mW/ $^{\circ}$ C; storage and operating temperature of -20 to 100 C; continuous forward current of 40 mA; peak pulse current (1 μ s pulse, 1% duty cycle) of 1 A; and peak reverse voltage of 5 V. The green LED has a typical forward voltage of 2 V, a reverse current of 1 μ A, a capacitance of 100 pF, a peak wave length of 560 nm, a spectral line half width of 30 nm and a luminous intensity of 1.5 mcd. Most of the yellow LED specs are the same. The only differences are the forward voltage of 1.9 V and the peak wavelength of 580 nm.

CIRCLE NO. 395

IR LEDs have peak transmission at 885 nm

International Rectifier Corp., 233 Kansas St., El Segundo, Calif. 90245. (213) 678-6281. From \$1.50 to \$3.20 (100-up).

The L-10 series of infrared LEDs is claimed by the manufacturer to offer an order of magnitude improvement in switching speeds, with the same or better output compared with other units on the market. The units are also more closely matched to the frequency response characteristics of fiber optics cable. Peak emission wavelength for the IR units is 885 nm. The 3-dB bandwidth of the L-10 diodes is 10 MHz. The first line of devices is housed in a TO-46 package. Other series, and a line of couplers, will be introduced later this year. Typical total radiation power output at 25 C and forward current of 100 mA is 1 mW for type LA 10-C; 2 mW for the LA 10-D; and 3 mW for the LA 10-E. Derating is 1.8 mW/ $^{\circ}$ C at $T_A = 25$ C, and 13 mW/ $^{\circ}$ C at $T_C = 25$ C. Maximum dc forward current at $T_C = 25$ C is 500 mA.

CIRCLE NO. 396

Vhf and uhf transistors deliver up to 50 W

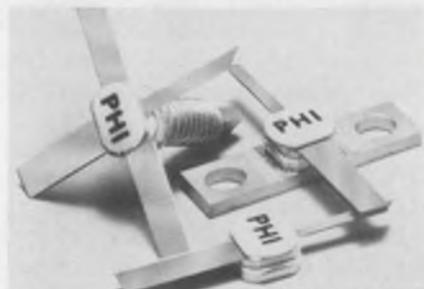


TRW Semiconductors, 14520 Aviation Blvd., Lawndale, Calif. 90260. (213) 679-4561. From \$13.85 (up to 999); stock.

A line of vhf and uhf transistors is rated for up to 50 W of output power. In the vhf range, devices are specified for output power of 4, 8, 15, 25 and 50 W. The PT9731, for example, has power output of 25 W and a gain of 10 dB. The PT9733, has power output of 50 W and a gain of 7 dB. In the uhf range, devices are specified for 4, 10, 20 and 30 W. As an example, the PT9704A is equivalent in performance to 2N-6105A, with rated output power of 30 W and gain of 7.8 dB.

CIRCLE NO. 397

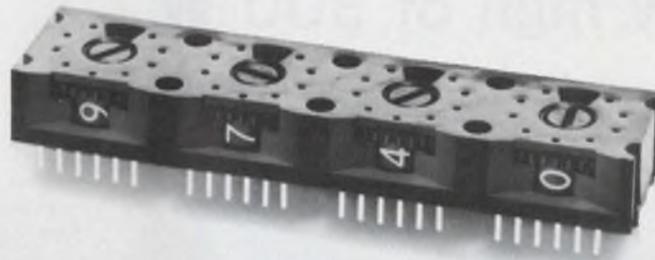
3-GHz transistor series delivers up to 6 W



Power Hybrids, Inc., 1742 Crenshaw Blvd., Torrance, Calif. 90501. (213) 320-6160. 1 to 99 prices: 3006, \$95; 3004, \$75; and 3001, \$51; stock.

A family of 3-GHz rf power transistors are all gold metalized and emitter ballasted. Included in this family are 6, 4 and 1 W transistors housed in hermetic ceramic-to-metal, flange, stud or pill packages. At 3 GHz Model PH3006 typically produces 6 W of rf power with 5.2 dB of gain and 33% efficiency at 28 V. at 2.3 GHz and 22 V the PH3006 will typically produce 8 W with 7 dB of gain.

CIRCLE NO. 398



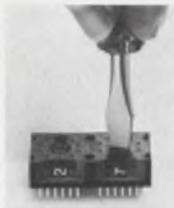
Introducing the programmable memory with visual readout.

We're talking about our unique Stripswitch,[™] the miniswitch for direct printed circuit board mounting or panel installation. This little marvel is molded of impervious Valox*, and is guaranteed for two years.



Other facts: it comes in one, two, three, four, five, and six station models (\$1.95 per station, standard); in a variety of codes, including decimal, BCD, Complimentary, Special Binary, and 1, 2, 3, and 4 pole.

On a custom basis we can do lots of other things to your Stripswitch. Like interconnections, markings and legends, color coding, stops, number of switch positions, additional stations, et cetera.



A 10¢ stamp will get you an immediate reply, or your finger on the dial of your telephone will get you instant answers. Call us collect. Or one of our distributors: G.S. Marshall, Hall-Mark, or Schweber.



Your sample awaits.

*Registered General Electric trademark

EECO

1441 East Chestnut Avenue
Santa Ana, California 92701
Phone 714/835-6000

35-MHz linear rf amp outputs new high of 300 W



Electronic Navigation Industries, 3000 Winston Rd. South, Rochester, N.Y. 14623. (716) 473-6900. \$4700; 60 days.

A new tubeless, class-A rf amplifier boasts the highest power output, 300 W, of any commercial unit in the frequency range of 0.6 to 35 MHz. In a pulsed or cw mode, Electronic Navigation Industries' Model A-300 power amplifier can output 500 W.

The new solid-state unit replaces bulkier, heavier, more costly tube types with far higher dissipation. And you can safely connect any load from short to open circuit.

Moreover the high power ratings don't require special electrical wiring. Any 115-V ac single-phase outlet that supplies 18 A can be used to provide primary power.

Similarly, special drive sources aren't needed. Because the 55-dB gain is flat to within ± 1 dB, any laboratory signal generator, synthesizer or transmitter can be used to drive the Model A-300.

Distortion is low, too. Second harmonics are at least 35 dB below the fundamental frequency at a

300-W output. Third harmonics are held to at least 25 dB below the fundamental at 250 W. Lower distortion results at reduced output powers.

Maximum transmitted power occurs when loads are 50 Ω —the impedance of the output as well as the input. However, any load can be connected to the output—or input—without amplifier failure or oscillation resulting. Reflected powers due to load mismatches are absorbed within the amplifier.

The unit can withstand more than 16-dB overdrive (input signal of 1 V rms) for all output load conditions. It has a maximum noise figure of 10 dB and operates over the 0-to-40-C temperature range. The compact unit measures 12-1/4 \times 17 \times 17 in. and weighs 88 lbs.

The impressive specifications for the Model A-300 result from the following: 16 custom rf-power transistors, a low-loss hybrid matrix network that combines the outputs of the transistors, and a cooling system that maintains the internal heat-sink plate to within 25 C of the outside air.

The rf transistors are said to be the largest devices for the frequency range. Also, the high-power custom chips use nichrome resistors deposited at the emitter sites to minimize hot spots and to even out the current distribution.

The manufacturer reports a full power burn-in of 100 hours for each device, and calculates MTBF (mean time between failure) for the complete amplifier to be more than 15,000 hours.

A special nickel-zinc ferrite material in the hybrid matrix contributes a maximum loss of only 0.1 dB at 35 MHz. Transformer hybrids using the ferrites are connected to form a 16-way summing network. The over-all maximum insertion loss is only 0.5 dB, or 12% of the power generated.

Maximum heat generation, totaling about 2000 W with 1400 W in the rf transistors, occurs under zero-signal conditions. To dissipate the power, amplifier modules are mounted on low-thermal-resistance aluminum plates. A cooling fan and corrugated fins on the heat-bearing plates combine to wash outside air over the modules.

CIRCLE NO. 250

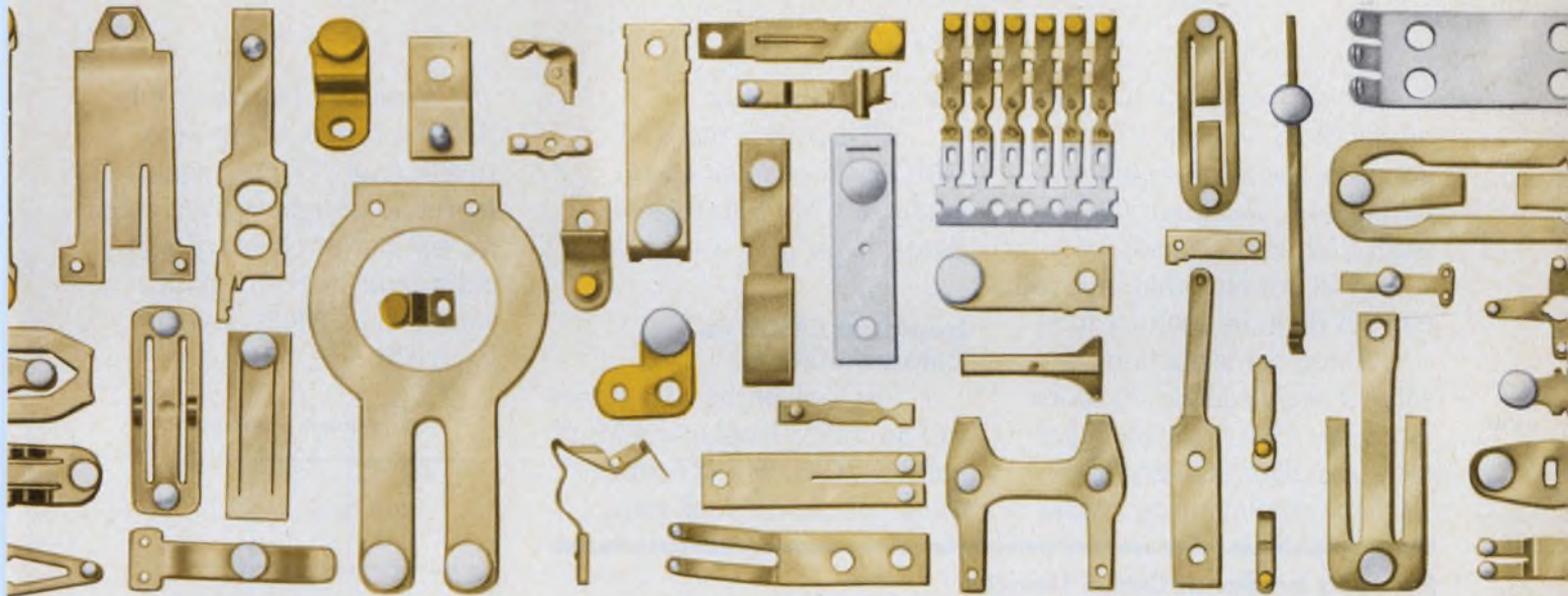
Transistor array works up to 500 MHz

RCA Solid State Div., Route 202, Somerville, N.J. 08876. (201) 722-3200. \$2.40 (1000).

A high-frequency npn transistor array has a frequency range up to 500 MHz. Called the CA3127E, the new array consists of five independent transistors, each with a gain-bandwidth product (f_T) in excess of 1 GHz. The unit comes in a 16-lead plastic DIP and operates over the full military temperature range of -55 to $+125$ C.

CIRCLE NO. 279

Which of these spring contacts can you get from Instrument Specialties?



All of 'em!

The number of different designs and combinations of springs and contacts is as boundless as your imagination. Yet, no matter what your requirements, chances are i/s can meet them. Specialists in the design and manufacture of beryllium copper springs, we can furnish these springs with your choice of many types of gold, silver or other precious metal contacts.

Perhaps you require large welded contacts in which precious metal is bonded to a layer of inexpensive material. Or, small welded contacts in which precious metal is welded directly to the spring. Or, riveted contacts offering close tolerances with either single or double headed contacts. Instrument Specialties has all of them!

One other thought: Sometimes, you may *think* you need welded or riveted contacts. However, our engineers may find that our CONTIP® bonding technique, or our gold selective plating process, can meet your needs at considerably less cost. If so, we'll tell you that, too.

Our latest catalog contains complete information on all types of i/s spring contacts. For your free copy, circle the Reader Service Card or write us at Dept. ED-51.



INSTRUMENT SPECIALTIES CO., INC.
Little Falls, New Jersey
Phone 201-256-3500

Isoplanar CMOS does what common CMOS can't. And now you can even get it in uncommon quantity.

Look, why settle for *common* CMOS when you can get all of the most popular functions in *Isoplanar* CMOS from Fairchild instead?

And not only that, but you can get them in quantity now.

Since our last ad on the subject, we've added 24 more Isoplanar CMOS device types for immediate delivery, with 19 more coming along before

the end of the year.

All together, that's 53 CMOS devices that leave ordinary CMOS far behind. And can put you way ahead.

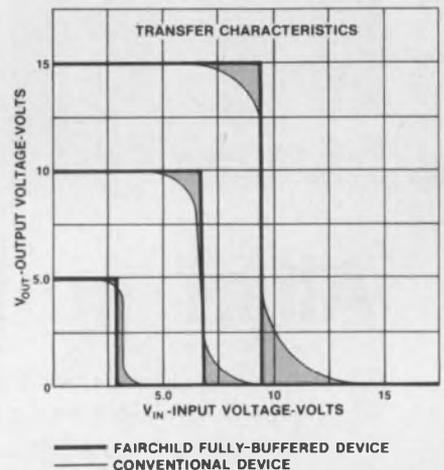
Isoplanar CMOS vs. Common CMOS.

Just look at the advantages of Fairchild's Isoplanar CMOS and you'll see why it's something you should look into.

Basically, Isoplanar fabrication reduces chip area substantially. Which means Fairchild designers have room for full buffer circuitry and extra built-in performance with every CMOS device. Even SSI.

Fairchild Isoplanar CMOS Devices.

| | | | |
|-----------------------|--|-------------------------------|---|
| <i>Available Now.</i> | | 340161 | 4-Bit Binary Asynchronous Reset Counter |
| 34001 | Quad 2-Input NOR Gate | 340162 | 4-Bit Decade Synchronous Reset Counter |
| 34002 | Dual 4-Input NOR Gate | 340163 | 4-Bit Binary Synchronous Reset Counter |
| 34011 | Quad 2-Input NAND Gate | 340194 | 4-Bit Right/Left Shift Register |
| 34012 | Dual 4-Input NAND Gate | 340195 | 4-Bit Shift Register |
| 34013 | Dual D Flip-Flop | <i>Available 4th Quarter.</i> | |
| 34015 | Dual 4-Stage Shift Register | 34014 | 8-Stage Parallel to Serial Shift Register |
| 34019 | Quad 2-Input Multiplexer | 34016 | Quad Bilateral Switch |
| 34020 | 14-Stage Timer | 34017 | Decade Sequencer |
| 34023 | Triple 3-Input NAND Gate | 34021 | 8-Stage Serial to Parallel Shift Register |
| 34025 | Triple 3-Input NOR Gate | 34024 | 7-Stage Binary Counter |
| 34027 | Dual JK Flip-Flop | 34029 | 4-Bit Binary/BCD Up/Down Counter |
| 34028 | 1-of-10 Decoder | 34035 | 4-Bit Parallel In/Out Shift Register |
| 34030 | Quad Exclusive OR Gate | 34068 | 8-Input NAND Gate |
| 34040 | 12-Stage Timer | 34099 | 8-Bit Addressable Latch |
| 34042 | Quad D Latch | 34104 | TTL-to-High-Level CMOS Converter |
| 34049 | Hex Inverting Buffer | 34518 | Dual 4-Bit Decade Counter |
| 34050 | Hex Non-Inverting Buffer | 34527 | BCD Rate Multiplier |
| 34051 | 8-Input Analog Multiplexer | 34555 | Dual 1-of-4 Decoder |
| 34052 | Differential 4-Input Analog Multiplexer | 34725 | 16 X 4 Bit RAM |
| 34069 | Hex Inverter | 340085 | 4-Bit Magnitude Comparator |
| 34071 | Quad 2-Input AND Gate | 340174 | Hex D Flip-Flop |
| 34077 | Quad Exclusive NOR Gate | 340175 | Quad D Flip-Flop |
| 34081 | Quad 2-Input OR Gate | 340192 | 4-Bit Up/Down Synchronous Decade Counter |
| 34085 | Dual 2-Input 2-Input AND/OR/Invert Gate | 340193 | 4-Bit Up/Down Synchronous Binary Counter |
| 34086 | Expandable 4-Wide 2-Input AND/OR/Invert Gate | | |
| 34512 | 8-Input Multiplexer | | |
| 340097 | Hex 3-State Non-Inverting Buffer | | |
| 340098 | Hex 3-State Inverting Buffer | | |
| 340160 | 4-Bit Decade Asynchronous Reset Counter | | |

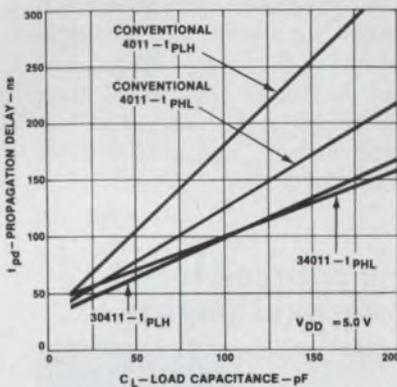


Because Fairchild devices are buffered for superior standardized performance, transfer characteristics are almost ideal.

And if you're a system designer, you'll really appreciate the benefits:

1. Highest guaranteed noise immunity in the industry.
2. Fully-standardized drive outputs for direct interface with low power TTL and low power Schottky TTL.
3. Pick your package: commercial grade plastic and military ceramic DIP is available now, commercial grade ceramic DIP in October.

4. Propagation delay that is less dependent on loading, for increased system speeds.



Fairchild's buffered CMOS gate has propagation delays which exhibit balanced T_{PLH} and T_{PHL} times and are less sensitive to load capacitance.

And the best part is, Fairchild's Isoplanar CMOS can replace any other 4000 series CMOS pin-for-pin.

So if you need production quantities of CMOS at competitive prices, call your Fairchild Distributor or Representative.

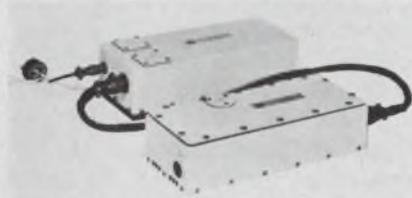
He's got all the facts on the kind of CMOS you need. Now.

**MADE IN
FAIRCHILD**

Semiconductor Components Group, Fairchild Camera & Instrument Corp., 464 Ellis St., Mountain View, CA 94040. Telephone (415) 962-5011. TWX: 910-379-6435.

MICROWAVES & LASERS

30-40-GHz front end has low noise figure

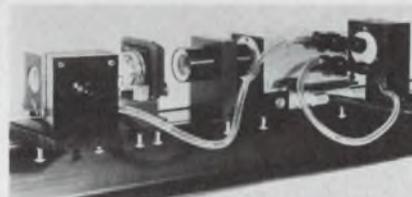


Alpha Industries, Inc., 400 Border St., East Boston, Mass. 02128. (617) 569-2110.

A low-noise radiometric receiver front-end operates in the 30-to-40-GHz range with noise figures of about 2.5 dB. The system consists of a degenerate parametric amplifier first stage followed by a low-noise, broadband mixer-amplifier section. A single Gunn oscillator source provides both pump and local-oscillator signals. The unit operates at 33.6 GHz with an overall gain of 70 dB over a bandwidth of 600 MHz. The system noise figure from input flange to i-f output port is typically 2.5 dB, or 226°K excess noise temperature. The package, which measures 3 × 5 × 12-in., operates over the temperature range of -40 to +140 F.

CIRCLE NO. 280

Fast laser yields narrow pulses

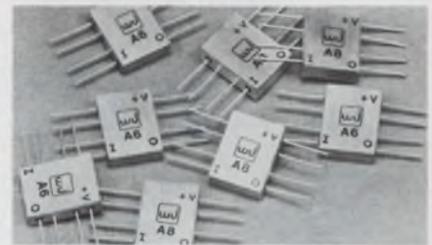


Korad Div. of Hadron, 2520 Colorado Ave., Santa Monica, Calif. 90404. (213) 829-3377.

Equipped with the KQS3 Q-Switch, the company's Model K15-QPTM laser system can achieve narrow pulse widths in the 3-to-5-ns range at high pulse-repetition rates. Pulse peak power is 60 MW minimum at a repetition rate of 60 ppm. The system consists of cabinet, mounting rail, and water-to-water cooling system. The cooler regulates the temperature of the polarizer and mirror mount to assure long-term stability in laser output at 60 ppm.

CIRCLE NO. 281

Flatbacks hold cascadable amps

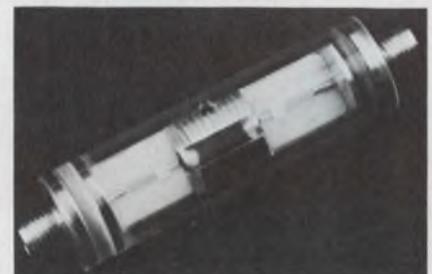


Watkins-Johnson Co., 3333 Hillview Ave., Palo Alto, Calif. 94304 (415) 493-4141. \$99 to \$109 (1-9); stock.

Now in 0.5 × 0.38-in. flatbacks, with a low profile of 0.125-in. maximum, are the company's cascadable amplifiers previously available only in TO-8 packages. The amplifiers have a frequency range of 5 to 500 MHz with typically 14.5 dB of gain over the two-decade bandwidth. Noise figure is typically 4 dB for the WJ-A6 and 5 dB for the WJ-A8. Output power is +9 dBm for the WJ-A6 and +14 dBm for the WJ-A8. All units are specified over the -54 to 71 C operating temperature range.

CIRCLE NO. 282

Acoustic component line expands

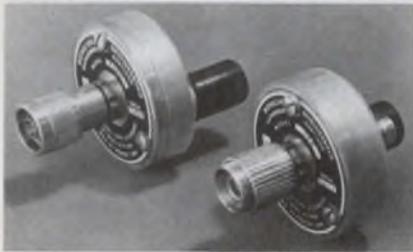


Thomson-CSF, Electron Tube Group, 8 rue Chasseloup-Laubat, 75737 Paris Cedex 15, France.

The company's line of acoustic-wave components expands with bulk-wave delay lines that reach to X-band, tapped SAW delay lines for generating and detecting programmable codes, and a family of Love-wave dispersive delay lines with 80% frequency excursion at up to 500 MHz. The new soldered-transducer X-band models have insertion losses of 10 dB, for 1-μs delay (plus 15 dB per transducer), relative passbands of 20% and time delays of from 0.5 to 2 μs. The Love-wave delay lines possess a very wide passband of 80% and an insertion loss of 40 dB.

CIRCLE NO. 283

Barretter mounts span 18-GHz range

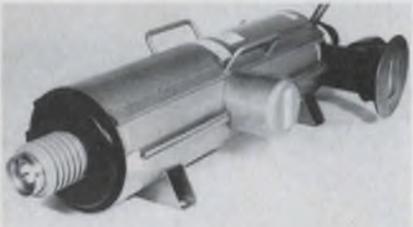


Weinschel Engineering Co., Box 577, Gaithersburg, Md. 20760. (301) 948-3434.

Barretter mounts with internal noise suppressors are available for the following frequency ranges: 50 MHz to 12.4 GHz—Models 926 and 928; and 50 MHz to 18 GHz—Models 927 and 929. These mounts feature a maximum SWR of 1.45 to 11 GHz, 1.75 to 12.4 GHz and 2.0 to 18 GHz. The Models 926 and 927 are furnished with Type N male or female connector, and the Models 928 and 929 are available with 7-mm precision connector.

CIRCLE NO. 284

S-band TWT delivers 160 kW peak



Thomson-CSF, Groupement Tubes Electroniques, 8, rue Chasseloup-Laubat, 75737 Paris Cedex 15, France

Incorporating a novel slow-wave structure, a high-power radar TWT, the TOP 1304, delivers 160 kW of peak output power—2 kW average—with a gain of at least 30 dB, anywhere in its S-band operating range of 2.85 to 3.15 GHz. With collector depression, its typical electrical efficiency—dc to rf—is 35%. To avoid the short-comings associated with coupled-cavity or conventional ringbar slow-wave structures, a solid copper stub-supported ring structure is used. It enables high peak-power operation with no danger of arcing, while lossy coatings suppress multimodes.

CIRCLE NO. 285

Wideband detector maintains sensitivity

Marconi Instruments Ltd., Sanders Div., Gunnels Wood Rd., Stevenage, England.

A wideband detector, Type 6160, operates in the frequency range of 10 MHz to 18 GHz. When loaded with a high impedance, the unit maintains its square-law characteristic to within ± 0.5 dB up to 40 mV output. As a result the sensitivity is at least 400 μ V output per microwatt input. Also available are factory-matched pairs.

CIRCLE NO. 286

Rugged SMA termination has small size



Midwest Microwave, 3800 Packard Rd., Ann Arbor, Mich. 48104. (313) 971-1992. \$17.50 (1-9); stock to 15 days.

A broadband SMA termination, covering the frequency range of dc to 18 GHz, is only 0.39-in. long. Moreover, the unit meets environmental specifications of MIL-STD-202. The Model 4444 is fabricated in stainless steel, and it has a maximum VSWR of $1.05 + 0.008 f$, where f is in gigahertz.

CIRCLE NO. 287

4-GHz tunable osc outputs 0.5 W

Frequency Sources, Inc., 166 Middlesex St., North Chelmsford, Mass. 01863. (617) 251-8747.

The Model FS-2148-1 cavity-stabilized oscillator can mechanically tune any 100-MHz band from 3 to 4 GHz at power levels to 500 mW. The afc option allows voltage tuning across $\pm 0.1\%$ of the band at typically 1 MHz/V and tuning rates to 10 MHz. FM noise is -90 dBc/Hz at 100 kHz from the carrier. Thermal frequency drift from 0 to 60 C is $\pm 0.15\%$. Operation into all phases of a 1.5:1 mismatch is typically ± 1 -dB power variation and ± 1 -MHz frequency pulling. Harmonic rejection is -35 dBc and all nonharmonics are greater than -60 dBc down.

CIRCLE NO. 288

Laser systems offer choice of powers



C W Radiation, Inc., 111 Ortega Ave., Mountain View, Calif. 94040. (415) 969-9482. \$190 up.

The LS Series of HeNe laser systems offers a broad selection of output powers: 1 through 5, 7, 10 and 15 mW. All laser systems have a regulated power supply for a $\pm 10\%$ change in ac-line variations. A precision current-regulated power supply is also available as an option to control the plasma current to better than 0.1% over a wide range of ac-line variations. Optional outputs are available for 1152 and 3391 nm.

CIRCLE NO. 289

Noise sources turn to mm wavelengths



Signalite, Div. of General Instrument Corp., 1933 Heck Ave., Neptune, N.J. 07753. (201) 775-2490. About \$1000.

Noise sources come in nine waveguide sizes covering the frequency range of 18 to 170 GHz. Low VSWR is obtained by use of a 7-degree insertion angle for the alumina ceramic noise tube and by use of a higher than normal pressure of argon gas along with a Kr85 additive. Units are furnished as transition type or with loads. Insertion losses in transition types are approximately 2 dB cold and 15 dB hot. Test data reveal a maximum cold VSWR of 1.30 and a hot VSWR of 1.20. ENR is 15.0 dB at 18 GHz and 13.6 dB at 90 GHz.

CIRCLE NO. 290

Announcing...

now there are more than 300 ways to improve your harnessing production with the **TY-RAP®** system
(if you need another way... we'll make it)

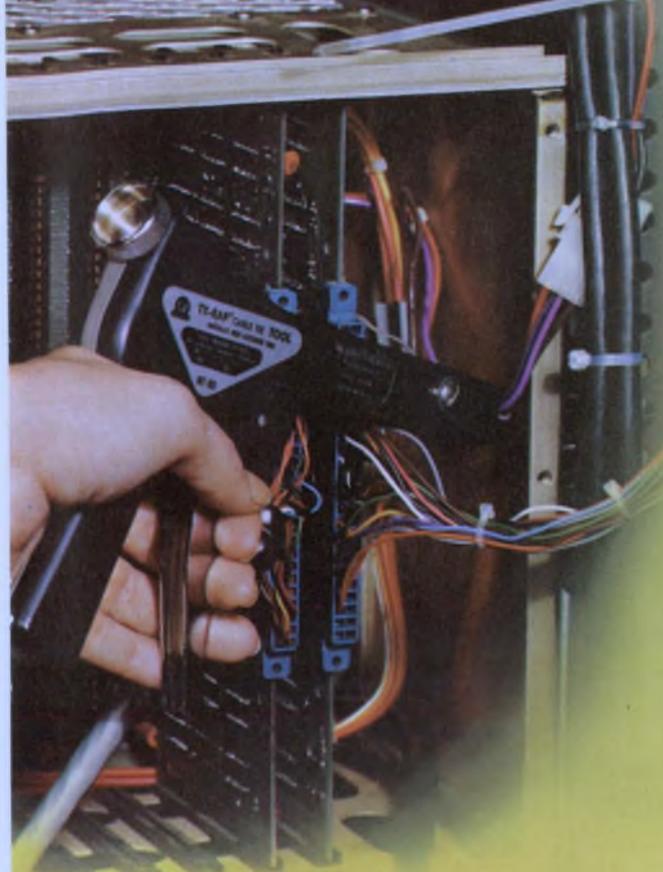
Anytime you have the opportunity to pick and choose exactly what you want, you increase design flexibility... you make a better buy. The TY-RAP line lets you do just that. You buy only what you need... without waste. There is a tie for just about every tying requirement, from 1/16" to 9" dia. Best of all, uniform reliability and most efficient production of your design is assured through the wide selection of tying aids and tools.

A major new benefit of the TY-RAP system is our national network of tool service centers to insure your continued high level of production efficiency.

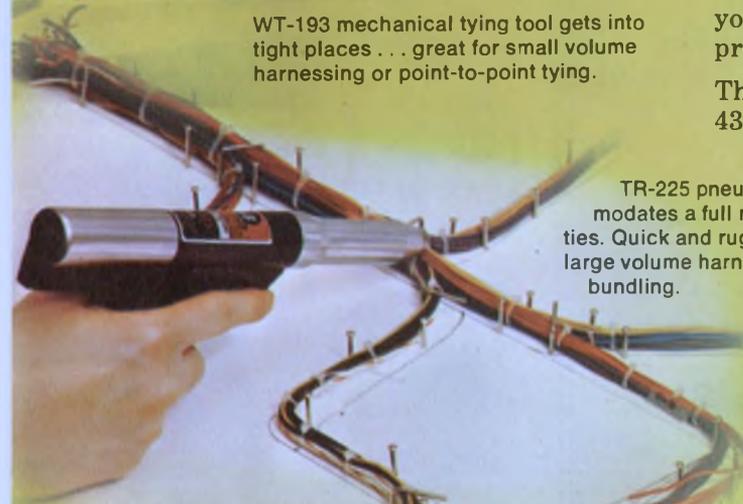
If you are not a user of TY-RAP products, perhaps our evaluation service for cost reduction techniques will help you. Call us, we'll be pleased to conduct a study of your production operation.

Thomas & Betts Co., Elizabeth, N. J. 07207. (201) 354-4321. In Canada, Thomas & Betts Ltd., P.Q.

Circle 69 on Reader-Service Card



WT-193 mechanical tying tool gets into tight places... great for small volume harnessing or point-to-point tying.



TR-225 pneumatic tool accommodates a full range of TY-RAP ties. Quick and rugged... ideal for large volume harnessing and cable bundling.



New TR-300 automatic tool applies ties in 8/10 of a second. Counterbalance system makes it a "feather-weight".

1/16" to 4"

1/8" to 3 1/2"

1/16" to 3"

1/16" to 1 1/4"

1/16" to 1 3/4"

1/16" to 2"

1/16" to 5/8"

Sold Only Through Authorized T&B Distributors



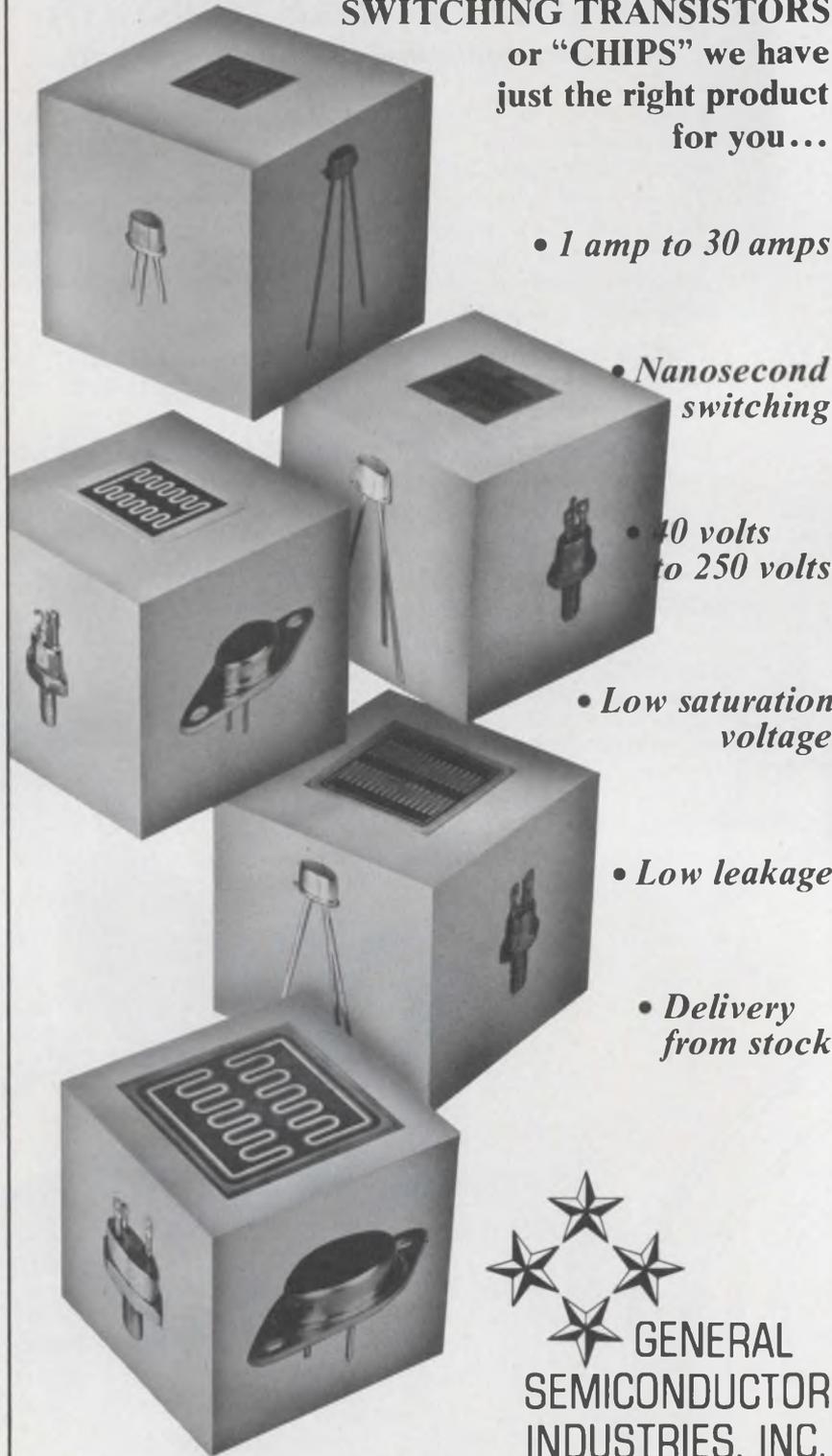
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voltage

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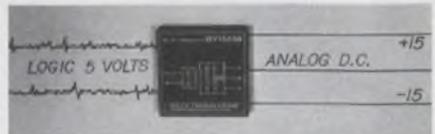


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POWER SOURCES

Dc/dc converter outputs
only 1 mV of noise



Data Translation, Inc., 109 Concord St., Framingham, Mass. 01701. (617) 879-3595. \$109 ea.; 2 wks.

The DT15150 dc/dc converter is designed to be compatible with the company's recently announced DATAX series of data-acquisition modules. The DT15150 converts 5-V to noise-free ± 15 -V power, capable of supplying 150 mA from each side of the 15 V. The unit offers multiple input/output isolation of 80-pF effective capacitance, 1000-M Ω resistance, and a breakdown voltage in excess of 500 V. Efficiency is 65% min. at ± 150 mA. Output noise voltage is a low 1 mV rms, an equivalent value of of less than 1/2 LSB in a 12-bit data-acquisition system.

CIRCLE NO. 291

Line-voltage regulators
work at 98% efficiency



Topaz Electronics, 3855 Ruffin Rd., San Diego, Calif. 92123. (714) 279-0831. Start at \$245.

These ac line regulators are offered in power ratings from 500 VA to 24 kVA, with outputs of 115 to 240 V ac. They operate from 47 to 63 Hz. And they have a minimum efficiency of 98%. Response time is less than 1 cycle. The output adds less than 1% total harmonic distortion. Audible noise is typically 2 dB. These compact units are designed for brownout conditions, when input voltage may drop as much as 30%.

CIRCLE NO. 292

Modular supplies offered for PC boards



ACDC Electronics, Oceanside Industrial Center, Oceanside, Calif. 92054. (714) 757-1880. Start at \$25; stock.

PCM series of miniature, encapsulated plug-in power supplies is designed for PC board mounting for op amps, d/a and a/d converters and nonlinear circuits. They come with built-in overload and short-circuit protection. For logic circuits, all 5-V models have built-in overvoltage protection. There are 13 models in this new series. Single output models range from 5 V at 500 mA to 24 V at 50 mA. Dual output models are available from ± 12 V at 120 to 400 mA and ± 15 V at 65 to 350 mA.

CIRCLE NO. 293

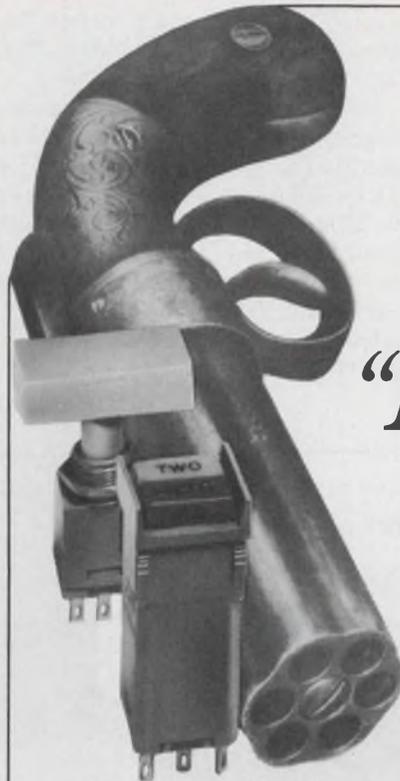
CRT supply gives more than 50% efficiency



LectroLogic Inc., P.O. Box 3727A, Milwaukee, Wis. 53217. (414) 351-1070. \$750 (1-4); 16 wks.

H Series CRT high-voltage dc/dc power supplies feature output voltages from 1 to 20 kV dc, up to 10-W output, efficiency greater than 50%, -55 to $+71$ C operating temperature range, and compact size—approximately 19 cu. in. Other features include remote programmability, current limiting and short-circuit protection, remote high-voltage disable in response to logic +5 V dc, and reverse polarity and input surge-voltage protection.

CIRCLE NO. 294



Before you order switchlights, we challenge you to compare our low cost

"Persuader Line"

feature for feature with other leading brands

We're the kind of firm that believes in more than one gun barrel and plenty of ammunition. So when you add our familiar S410* series to our new S190* series, you'll find we have a very convincing line of general purpose switchlights indeed. It's "The Persuader"—the line we invite you to compare for low cost, quality and versatility with that of any other manufacturer. Just check the list below, then get in touch with your local distributor for exact specifications. And we're easy to find . . . located in major cities world wide.

| Standard Features | Clare-Pendar "Persuader" | Micro | Diallight | Other |
|--------------------------|--------------------------|-------|-----------|-------|
| 1. Low Cost | YES | | | |
| 2. Distributor Stock | YES | | | |
| 3. U.L. Listed | YES | | | |
| 4. 2 Form C | YES | | | |
| 5. Wiping Contacts | YES | | | |
| 6. Snap Action Contacts | YES | | | |
| 7. 10 amp Rated | YES | | | |
| 8. 2 amp Rated | YES | | | |
| 9. 100,000 Cycle Life | YES | | | |
| 10. 6 Lens Shapes | YES | | | |
| 11. Split Lens Displays | YES | | | |
| 12. Solid/Proj. Displays | YES | | | |
| 13. 5 Adapter Shapes | YES | | | |
| 14. Barrier Adapters | YES | | | |
| 15. Snap-In Mount | YES | | | |
| 16. Rear Panel Mount | YES | | | |
| 17. Gang Frame Mount | YES | | | |
| 18. Quick Connect Trmls. | YES | | | |
| 19. Engraved Legends | YES | | | |
| 20. Alt. Remain-In | YES | | | |
| 21. Mom./Alt./Indicator | YES | | | |

*S190 \$1.62 in quantities of 1000

*S410 \$2.53 in quantities of 1000

CLARE-PENDAR

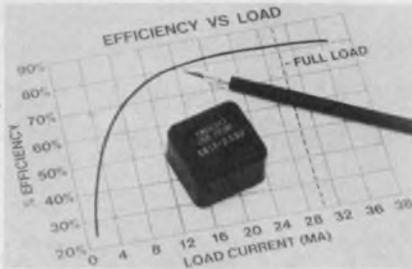


CLARE-PENDAR
a GENERAL INSTRUMENT CO.
Box 785, Post Falls, Idaho 83854
(208) 773-4541

INFORMATION RETRIEVAL NUMBER 94

POWER SOURCES

Dc/dc converters drive high-voltage displays



Endicott Coil Co., 24 Charlotte St., Binghamton, N.Y. 13905. (607) 797-1263. \$13.50 (lm); 3-4 wks.

Series 800 is a dc-to-dc power supply for matrix-type readout displays, such as Burroughs' SELF-SCAN®. The series converts 9, 12 or 15 V to a nominal 250-V level at 30 mA. Size is 1.43 × 1.5 × 1.03 in. and weight is 40 g. Series 800 units are fully encapsulated and are designed for PC-board mounting.

CIRCLE NO. 295

Batteries hold charge for long periods



Wisco Div., ESB, Inc., 2510 Louisburg Rd., Raleigh, N.C. 27604. (919) 834-8465.

The Willard brand of Charge Retaining (CR) storage batteries holds a charge over long periods and still provides uniform voltage. Pure lead grids hold stand loss to only 15% of capacity per year at 80 F. At current rates that discharge batteries over a period of 90 days or longer, voltage is practically linear throughout 90% of the discharge. The batteries are available in 2 and 6 V, with up to 1200 A-h rated capacity.

CIRCLE NO. 296

When a standard switch won't do...
take advantage of P&B's free engineering program



After you've seen the standard switches and still can't find what you need, call us. P&B offers unique design and manufacturing capabilities in actuator assemblies, push buttons and pneumatic actuated units. From dry circuit to 25 amps. And, there's no charge for reasonable initial engineering services and samples.

Besides the more common actuating assemblies, special P&B designs can include push on/push off, rocker maintain, rocker momentary, bar lever, air cylinders, relays, rotary, and many more. P&B was the first to qualify with U/L TV rated switches.

For help with your switch problem, call your P&B representative or write Potter & Brumfield Division AMF, Princeton, Indiana 47670. 812.385.5251.



Open-frame supplies offer 1% stability

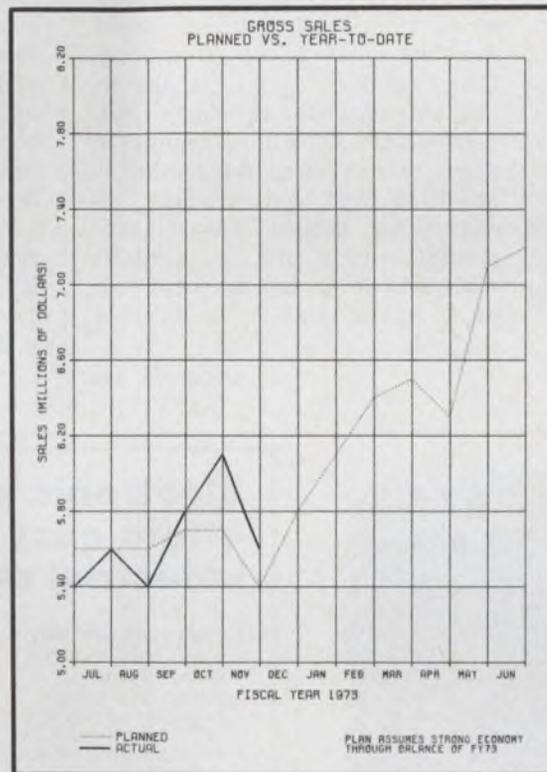


Deltron Inc., Wissahickon Ave., North Wales, Pa. 19454. (215) 699-9261. Start at \$28 (10); stock.

EPS Series of open-frame power supplies incorporates an additional frequency-compensating network for improved transient response and better stability under varying load changes. Line and load regulation is 0.5% with 0.1% available as an option. Stability is typically 1% for 8 h after warmup. Cut-back current limiting protects the supply from overloads. These units are available in 15, 18, and 24-W models. Size is 4 × 4-1/2 × 2-1/2-in. and weight is 2 pounds.

CIRCLE NO. 297

After your 360/370 massages the information, a Gould Plotmaster can draw you a picture within 3 seconds.



If alphanumeric information is what you want, a Gould Plotmaster can print it for you. At speeds up to 3000 lines per minute. But there are times when alphanumeric listings are just too much. Too much paper to handle, too tough to read, too difficult to digest.

And it's at times like these that a Gould Plotmaster can draw you a picture. A line chart, a bar chart, a pie chart, a graph. A picture that tells your story at a glance.

Employing high-speed electrostatic printer/plotters, Gould Plotmaster Systems give you power and versatility for both on-line and off-line operation. And they're designed to run on any IBM System/360 or 370 operating under DOS or OS, real or virtual.

Easy-to-use software packages help our Plotmaster Systems do the whole job. There's our DISPLAY package that provides even non-programmers with the capability of easily generating line, bar and pie charts. And there's our PLOT package which, due to the speed

and flexibility of our printer/plotters, lets you do background grids, variable line weights, automatic stripping, text annotation, and allows you to erase previously programmed line segments.

All software is written in assembler language to conserve

internal storage and make efficient use of your main computer — yet each subroutine package is callable from Fortran for ease of operation and efficiency in preparing input.

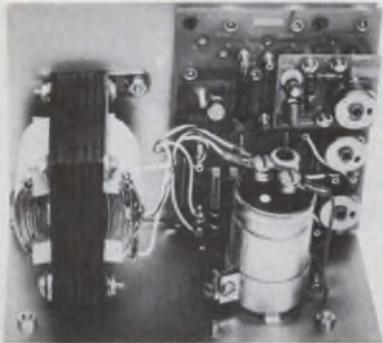
As for the hardware itself, our Plotmaster Systems can provide on-line/off-line operation, paper widths up to 22 inches, resolution up to 100 dots per inch, output speeds up to 7 inches per second. And, of course, a printing capability, as well.

Get all the facts on Plotmaster Systems from Gould Inc., Instrument Systems Division, 20 Ossipee Road, Newton, Mass. 02164 U.S.A. or Kouterveldstraat 13, B 1920 Diegem, Belgium.

 **GOULD**

POWER SOURCES

Open-frame supplies give 2-year warranty

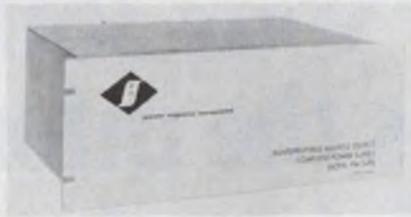


NJE Corp., P.O. Box 50, Culver Rd., Dayton, N.J. 08810. (201) 329-4611. \$85 (5V at 9 A).

These open-frame supplies come with a two-year warranty on labor and material. Other features include computer-grade electrolytic capacitors; plug-in IC regulator; and rack-adaptor capability with various front and rear-panel options. The units come in 12 models from 5 V at 3 A to 5 V at 15 A with 12, 15, and 24-V versions also included.

CIRCLE NO. 298

800-W computer supply works under blackouts



Pioneer Magnetics, 1745 Berkeley St., Santa Monica, Calif. 90404. (213) 829-3305. \$897 (100 quan.); 150 days, ARO.

Designated the Model PM 2478, this 800-W uninterruptible, multiple-output computer power supply can operate over an input range of 90 to 130 V ac, and with power dropouts up to 20 ms in duration. Under blackout conditions, the device automatically switches to battery operation. Upon resumption of power, the system returns to ac operation and automatically recharges the battery. The standard multiple-output unit is available with 5 V dc at 100 A, -15 V dc at 5 A and ± 15 V dc at 15-A channels.

CIRCLE NO. 299

Ac power source delivers 250 VA



California Instruments, 5150 Convoy St., San Diego, Calif. 92111. (714) 279-8620. \$690; 30 days.

The Model 251TC is a 250 VA, single-phase power source. Full power is obtained from 45 to 5000 Hz, and its two output voltage taps of 0 to 67.5 and 0 to 135 V make it convenient to stack two units to achieve 500 VA of single-phase power. Total harmonic distortion is less than 0.9% over the frequency range with an amplitude stability to $\pm 0.25\%$. Load regulation is $\pm 1\%$ over the frequency range while line regulation is $\pm 0.25\%$ of full output for a $\pm 10\%$ line change.

CIRCLE NO. 300

PORTABLE SPECTRUM ANALYZER 4-1000 MHz



Texscan's AL-51 spectrum analyzer is a general purpose, high reliability instrument designed for making rapid signal strength measurements in the 4 to 1000 MHz frequency range. The unit is completely portable, can operate off its own internal rechargeable batteries or from an external 12 to 20 volt source, including vehicle 12V battery. Weighing less than 25 pounds, it is provided with a protective cover and a carrying handle for simple field operation in adverse environments.

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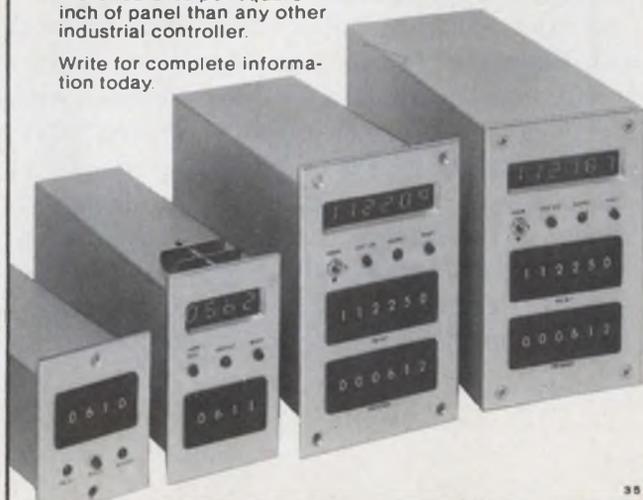
INFORMATION RETRIEVAL NUMBER 97

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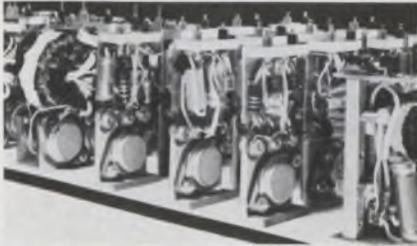


35

INFORMATION RETRIEVAL NUMBER 98

ELECTRONIC DESIGN 20, September 27, 1974

Submodules offer 1200 supply configurations



Arnold Magnetics, 11520 W. Jefferson Blvd., Culver City, Calif. 90230. (513) 870-7014. \$62 up; 10 days.

Thin-Mod submodules provide over 1200 possible configurations of high-efficiency ac/dc and dc/dc regulated power converters. Each completed system is provided in a tested and encapsulated miniaturized package. Available inputs include 115 V ac, 50 to 500 Hz and 12, 28, 48 and 115 V dc. Output submodules are offered from 4.2 to 300 V dc, and feature line and load regulation to 0.1%. Up to 120 W per output can be obtained and efficiencies run as high as 85%. All regulator submodules are short-circuit protected.

CIRCLE NO. 301

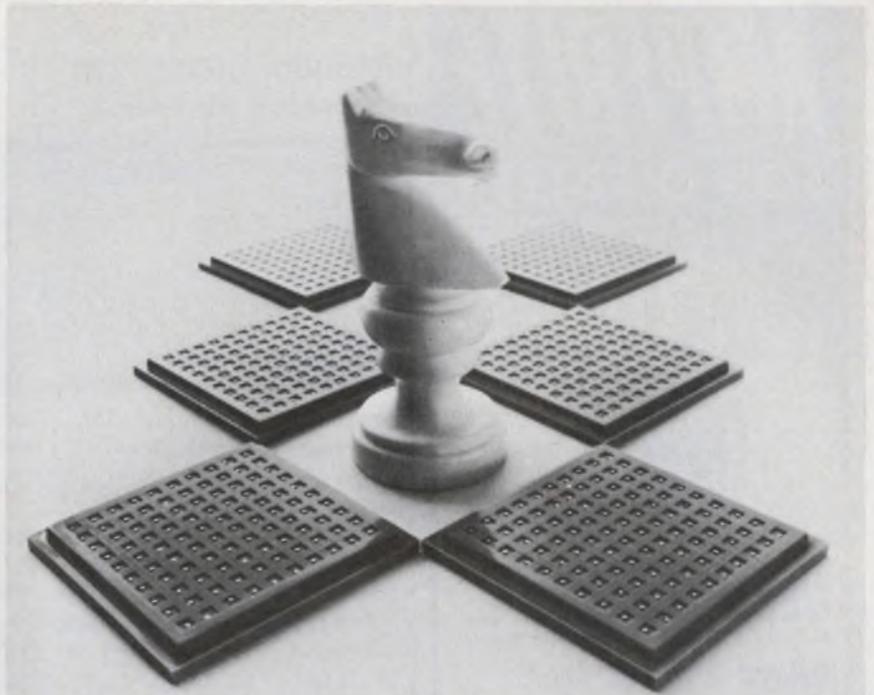
Unregulated converters offer 80% efficiency



Semiconductor Circuits, Inc., 306 River St., Haverhill, Mass. 01830. (617) 373-9104. \$25.95 to \$30.95; stock to 8 wks.

Unregulated miniature dc/dc converters come in encapsulated modules and deliver 900 mW. Packaged in a 1-1/8 x 1-1/8 x 1/2-in. module, these devices convert 5, 6, 12, 24 or 28 V to an isolated 5, 12, ± 12 or ± 15 -V output at efficiencies of more than 80%. Regulation is $\pm 3\%$ from 10% load to full load.

CIRCLE NO. 302



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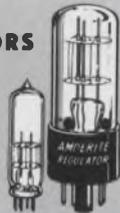
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Telephone: 201 UNION 4-9503

In Canada: Atlas Electronics, Ltd.,
50 Wingold Ave., Toronto 19

INSTRUMENTATION

Pseudorandom gen operates at 1 GHz

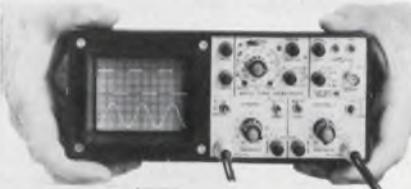


Tau-Tron, 11 Esquire Rd., North Billerica, Mass. 01862. (617) 667-3874. MN-2, starts at \$11,000; 120 days.

The MN-2 Pseudorandom Data Generator and the MX-2 Data Sampler operate at 1 GHz. The MN-2 produces a pseudorandom sequence at bit rates from 200 MHz to 1 GHz. In actual applications, the MN-2 output is used to excite the system under test. Output from the system under test drives the MX-2, which samples the received data. The sampling process lowers the data rate to a 300-MHz maximum rate, but preserves the pseudorandom character of the signal. To measure errors in the sampled signal requires a standard S-1100A PCM Receiver/Counter.

CIRCLE NO. 303

Dual-trace mini scope brags advanced trigger



Vu-Data Corp., 7170 Convoy Ct., San Diego, Calif. 92111. (714) 279-6572. \$1095; stock to 30 days.

Model PS940A Mini-Portable oscilloscope offers computerized triggering: a logic circuit eliminates front-panel adjustments to achieve a stable trace display. Once power has been turned on and the input signals are connected, the operator is guaranteed a stable, dual-trace presentation at all times. Furthermore, vertical position adjustments can be made without losing sync. Other specs include a 20-MHz bw, 10 mV/div sensitivity, a built-in delay line and full dual-trace switching.

CIRCLE NO. 304

2-MHz pulse gen sells for just \$159

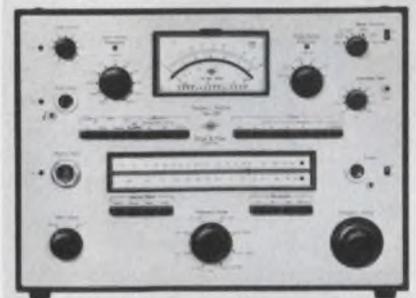


Interdesign, Inc., 1255 Reamwood Ave., Sunnyvale, Calif. 94086. (408) 734-8666. \$159.

In the Model 1101 pulse generator, 85% of all components are contained in a single custom IC. The unit is battery operated, so it can be isolated from ground. The batteries provide 40 h of operating time and are rechargeable in 10 h with the included charger. Frequency range is more than seven decades: 0.1 Hz to 2 MHz. Fall time is 10 ns and rise time is 50 ns. Pulse width is continuously variable from 0 to 100% of the period. Output voltage is adjustable from 0 to 5 V.

CIRCLE NO. 305

Event recorder captures transients



B & K Instruments, 5111 W. 164th St., Cleveland, Ohio 44142 (216) 267-4800. \$8397; 30 days.

Digital Event Recorder, Model 7502, contains an 8-bit a/d converter, an 8-bit shift register with 40-k words and a d/a converter for analog data recovery. The capacity can be increased in 2-k word increments to 10-k words. Variable sampling and playback rates provide speed scaling. A transient trigger provides automatic data capture. Multiple units may be interlocked for synchronous multichannel recordings. Built-in anti-aliasing filters simplify application. Frequency range is dc to 50 kHz.

CIRCLE NO. 306

Centralab perspectives

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Electronics Division

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THIS CATALOG, available from your Distributor, gives complete specs, dimensional drawings, circuit diagrams and easy ordering instructions. Ask for it. You'll find the answers to your design and delivery problems.

THE SELECTASHAFT SYSTEM consists of preassembled standard rotary switches and a series of individual exact shafts. Factory trained Distributors, using specially designed equipment, assemble these parts to meet custom requirements. You choose from three shaft styles—.250" plain round, .218" and .156" flat. There are 24 shaft lengths, .687" to 2.375" and 24 shaft flat angles in 15° increments. You select from 92 subminiature and miniature switches—1", 1.325" and 1.500" diameters; diallyl phthalate, phenolic or ceramic sections. You have your choice of knobs, dial plates, index assemblies and hardware. You get the design freedom of over 100,000 possible custom switch combinations.

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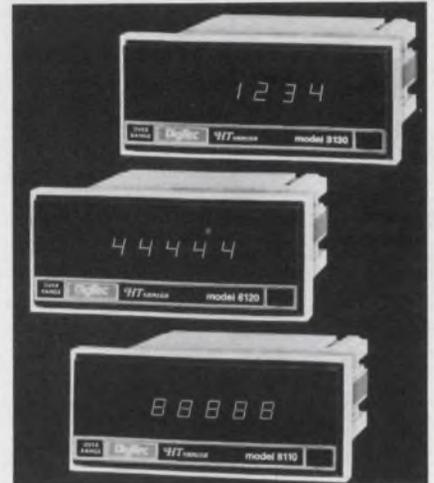
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5988

INSTRUMENTATION

Panel counter-timers display any units



United Systems Corp., 918 Woodley Rd., Dayton, Ohio 45403. (513) 254-6251. \$275; stock.

Consisting of three different models, the 8100 Series of factoring counter-timers count, totalize or time and provide a direct or factored digital readout in the desired units of measure. Error producing and time-consuming calculations, interpolations or look-up tables are no longer required. The series uses an internal crystal clock and each model comes with a 5-digit, Monsanto MAN 72 LED display. The 8110 is a counter; the 8120 a totalizer; and the 8130 a timer.

CIRCLE NO. 307

pROM programmer checks 256-bit memories

Tycom Corp., 26 Just Rd., Fairfield, N.J. 07006. (201) 227-4141. \$895; 30 days.

A programming system, the PROVER I, is designed to program or verify 256-bit pROM ICs. The system can duplicate one master 82S23 or 8223 pROM and verify another simultaneously. It can also be used to program or verify in a single step, permitting the operator to program, verify, or modify information on a particular address within the circuit. Other characteristics of PROVER I include a duplicate time of 80 s, verify time of 80 s and indefinite data retention time.

CIRCLE NO. 308

Panel DVM measures only 9/16-in. thick



Nationwide Electronic Systems, 1536 Brandy Pkwy., Streamwood, Ill. 60103. (312) 289-8820. \$199 to \$225.

The Slimline digital voltmeter—only 9/16-in. thick—is a bipolar, 3-1/2-digit instrument with LED display and BCD outputs. The unit mounts flat on the front of the panel. Specs include 10 readings per second and an accuracy of better than $\pm 0.05\%$ of reading ± 1 count. The unit requires 5-V dc power.

CIRCLE NO. 309

Transistor tester handles 2.5 kV, 1 mA

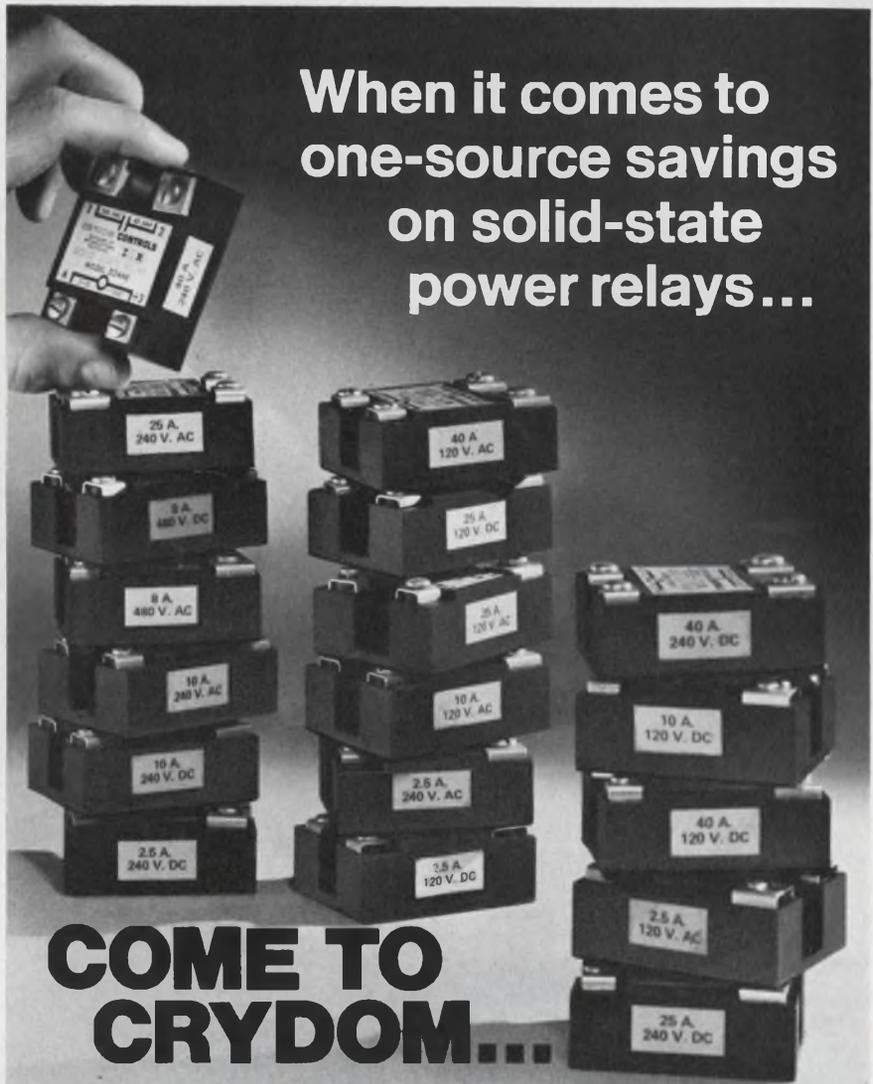


Teradyne, 183 Essex St., Boston, Mass. 02111. (617) 482-2700. \$15,000; 20 wks.

The T397 high-voltage transistor test instrument automatically tests npn transistors at voltages up to 2500 V. The unit can function independently or with the T347 computer-operated transistor test system. The T397 tests breakdown voltage and leakage current at up to 2500 V and 1 mA. Latching voltage and current gain can be tested at up to 1000 V and 100 mA. The internal memory of the instrument has a basic capacity for 32 tests.

CIRCLE NO. 310

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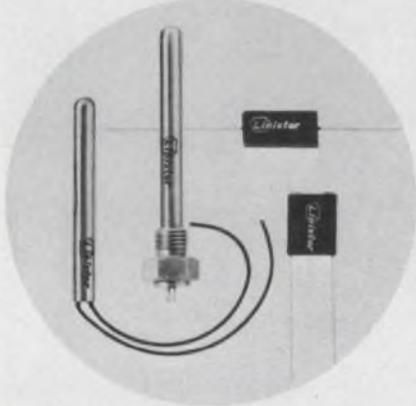
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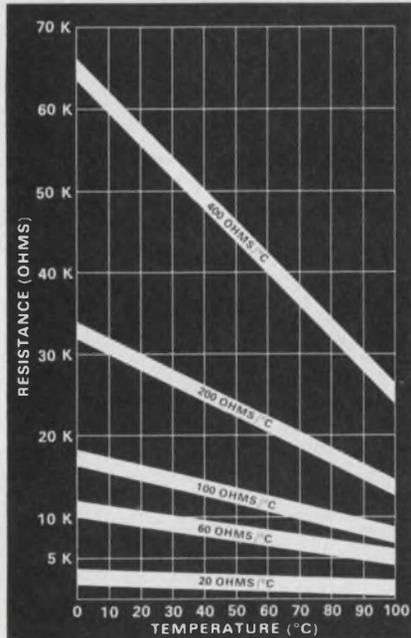
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INFORMATION RETRIEVAL NUMBER 104

INSTRUMENTATION

Unit determines semi parameters



Princeton Applied Research, Box 2565, Princeton, N.J. 08540. (609) 452-2111. \$2000; 90 days.

Model 410 C-V plotter determines semiconductor flat-band voltage, mobile ion concentration, surface-state concentration and can be used for carrier lifetime studies. The unit contains a 1-MHz capacitance meter, a variable dc voltage for bias stressing the sample and a ramp generator/programmer designed specifically for semiconductor applications. Completely integrated systems containing the Model 410 and accessory items such as an X-Y recorder, automatic temperature controller and hot chuck and various probe stations are available.

CIRCLE NO. 320

Logic tester fits in a suitcase



Fluke/Trendar, 500 Clyde Ave., Mountain View, Calif. 94040. (415) 965-0350. \$5975; 30 days.

TRENDAR 1000A is a high-speed, portable tester, self-contained in a suitcase and with the capability of one pass, go/no-go testing. The unit is a successor to the TRENDAR 1000. The 1000A accepts up to 60 output signals from the board to be tested and combines these into one digital bit stream which represents the "signature" of the UUT. The transition counter probe can now be placed on the output test point for a complete go/no-go test of the UUT in one pass.

CIRCLE NO. 321

\$245 buys 2-MHz function generator



Krohn-Hite Corp., 580 Massachusetts Ave., Cambridge, Mass. 02139. (617) 491-3211. \$245; stock.

Model 5800 function generator is a rep-stocked unit that provides sine, square and triangle waveforms at low cost. Features include: distortion less than 0.5%; frequency range of 0.2 Hz to 2 MHz; 50-Ω output with 15 V pk-pk and pushbutton operation. In addition, the 1000:1 tuning dial covers the entire audio range in one turn of the dial with no multiplier adjustments.

CIRCLE NO. 322

Multifunction counter offers remote control



Modular Devices, Inc., 1385 Lakeland Ave., Bohemia, N.Y. 11716. (516) 567-9620. \$725 with 0.0025% time base.

Model 2319FE multi-function counter offers remote control as standard. All functions are remotely or locally controlled by non-critical TTL/DTL signals. Parallel and serial data output is standard. Serial data format is either low or high-order bit first, field selectable. The unit, which mounts in a 19-in. rack, provides time-interval measurements of pulse width, period, B-A (to 100-ns resolution), frequency ratio E/B. Measurement averaging of ten consecutive readings increases time resolution to 10 ns.

CIRCLE NO. 323

**RF Microwattmeters—
Analog and Digital—
FS Sensitivities from
10 nW to 10 mW**



MODEL
42B/42BD

Boonton rf microwattmeters offer unrivalled sensitivity: 10 nW fs to 10 mW fs; from 200 kHz to 18 GHz, at highest stability ever attained at these sensitivities. Analog (42 B) or digital (42 BD) versions, both with linear DC outputs and logic-level programmability. BCD outputs are standard on digital version, auto-ranging and dB display (0.01 dB resolution) optional. Boonton Electronics, Parsippany, N.J. 07054

INFORMATION RETRIEVAL NUMBER 161

**Wide-Range
Programmable
Capacitance Meters**



MODEL
72B/72BD

Boonton analog (72B) and digital (72BD) provide rapid, accurate, 3-terminal and differential measurements, at 1 MHz, from 1 pF fs. Measures semiconductor-junction capacitance at low (15 mV) test level, with provision for external DC bias. Phase-sensitive detector measures accurately even at Q=1. Logic-level range programmability and fast-tracking DC output are ideal for ATE. Model 72BD has standard BCD output and autoranging. Boonton Electronics, Parsippany, N.J. 07054

INFORMATION RETRIEVAL NUMBER 162

**Direct Capacitance Bridge
0.00005 to 1000 pF
and 0.01 to 1000 mho**



MODEL
75D

Boonton Model 75D Direct Capacitance Bridge is designed for 1 MHz capacitance and loss measurement. Capacitance range, 0.00005 pF to 1000 pF; basic accuracy, 0.25%. Conductance range: 0.01 mho to 1,000 mho, basic accuracy, $\pm 5\%$. Internal bias from -6 V to +150 V. Adjustable test level, 1 mV to 250 mV, at 1 MHz. Two modes of operation allow either conventional capacitance and loss measurements or one-control balance for capacitance only. 3-terminal input configuration. Boonton Electronics, Parsippany, N.J. 07054

INFORMATION RETRIEVAL NUMBER 163

INFORMATION RETRIEVAL NUMBER 164 ►



**Why buy the
HP 8640B at \$5,400
when \$2,205 less
will get you Boonton's
competitive 102A
signal generator?**

Our Model 102A, at \$3,195, has everything you need for just about any AM/FM application — plus seven performance and convenience features you won't get in the HP \$5,400 design.

What did we leave out?

Phase-lock synchronization, for one (but our dc-coupled FM channel can be externally locked if you need better stability than our typical 4 ppm); and narrow-pulse modulation (belongs in a different class of generators).

What did we add?

Four different signal-generation techniques — for optimum performance in each band, from 4.3 to 520 MHz, without the usual compromises in noise, stability, or residual-distortion characteristics.

The most logical panel layout and convenient control setup you've ever seen. And a unique adjustable "feel" main drive mechanism for narrow-band receiver setting with ease — even without our electrical vernier.

Separate meters for modulation and output — no annoying auto-ranging or out-of-range annunciators . . . we don't need them.

15 minute warmup to typically

meet 10 ppm/10 minute stability — made possible by low internal dissipation (only 30 watts; no fan!)

Wider FM deviation at low carrier frequencies than any other design in this class (how does 2 MHz peak-to-peak grab you?)

A detected-AM-output option, to verify our negligible phase-shift for VHF-omni testing.

Versatile modulation features — like five internal frequencies, 30% and 100% AM scales, and true-peak-responding AM and FM metering.

All these performance pluses are coupled with low spurious and close-in noise, excellent low-frequency phase integrity, really effective leveling, a low and flat VSWR curve, accurate wide-range attenuation, high output power . . . all of it buttoned up tight for low leakage in a lightweight 30 pound package.

. . . and it's all yours for \$3,195. Get the full specs today — before you spend 70% more . . .

For complete data or a demonstration write or call Boonton Electronics Corp., Rt. 287 at Smith Road, Parsippany, N.J. 07054, (201) 887-5110.

BOONTON

ALPHA NUMERIC DOT MATRIX THERMAL PRINTHEADS ARE HERE!

DL 1100
10-DOT CHARACTER
PRINTER TO 18 c.p.s.

Features:

- Thick film — low cost
- Closely spaced square dots — high legibility

DM 1101
10 column
expandable
page printer
8 l.p.s.

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Gulton Industries Inc.
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MODULES & SUBASSEMBLIES

Adjust on and off times separately on timer

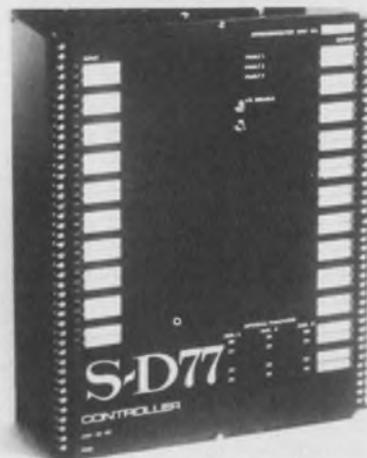


Metrix Manufacturing, 51 Sheffield St., Toronto, Ontario M6M3-E5. (416) 249-9153. \$85.

The Model CT/LR recycle timer has on and off times independently adjustable from 1 to 100 h or longer. It is specifically designed to drive an external relay or contactor. The unit is available for operation from 12 V dc or 115 V ac 50/60 Hz, and has a spst/NO solid state output rated at 1.5 A. Operation from other supply voltages can be provided on request. Each timer is equipped with two controls for independently adjusting the on and off times as a percentage of the maximum time periods. The CT/LR is available in a wide range of maximum on and off times. A LED indicator as to whether the CT/LR is in the on or off cycle is optionally available. The CT/LR can also be supplied with an extra set of output contacts to provide a direct reading of elapsed time on any external 1 mA meter. Repeat accuracy of the CT/LR is $\pm 1\%$ at 25 C, $\pm 0.2\%$ typical. Transient protection of the LSI timing circuit is 2 kV for 10 ms or 1.2 kV for 20 ms. Temperature coefficient is $\pm 0.15\%/^{\circ}\text{C}$ typical for time ranges up to 30 min. Reverse polarity protection is provided on dc models and the entire unit is housed in a plastic case with built-in screw-type terminal connections. The CT/LR is protected against moisture, chemicals and dust, and operates over the temperature range -10 to $+55$ C.

CIRCLE NO. 324

Programmable controller made for simple gear



Struthers-Dunn, Inc., Pitman, N.J. 08071. (609) 589-7500. Under \$800.

The S-D77 programmable logic controller (PLC) is designed for the "low end" of the industrial control market. It offers two design features not found in most other controllers. The multiple-slaving capability enables one PLC to directly communicate with another, thus allowing for distributed control functions on transfer lines, printing presses, etc. The vertical, slim-profile configuration enables the unit to fit in a standard 8-in. deep relay cabinet. The S-D77 is housed in a $12 \times 16.125 \times 7$ in. package and has 20 inputs, 12 outputs, 32 internal storage (scratch pad) locations and four timers. Included in the standard unit is all necessary hardware to process 120 V ac/dc input signals and deliver output signals in 4 ms. Also included as standard are fault monitoring for bus and scanner, I/O terminals and status lights, isolated inputs and fused outputs and a reprogrammable memory.

CIRCLE NO. 325

Temperature stabilizer preset for 65 C

Jermyn, 712 Montgomery St., San Francisco, Calif. 94111. (415) 362-7431. Stock.

The 4ST2-4 temperature stabilizer can control the temperature of TO-5 size devices at 65 C. The stabilizer is suitable for ICs where the lead lengths are restricted to 0.5 in. Power requirements are 24 V ac/dc, 0.6 W (at 25 C ambient). Maximum warm-up time from -55 C is 3 min.

CIRCLE NO. 326



NEW Simpson 360 Digital VOM

Handy and rugged enough for the field — accurate and versatile enough for the lab. And its simplicity and readability make it perfect for production line testing.



- 3-1/2" digit, non-blinking, auto-polarity 0.33" LED display
- AC line or battery operation
- 29 popular ranges including "low-power" ohms
- Analog indicator for scanning nulls and peaks
- Analog output signal permits interfacing with recorders and other instruments
- Up-down integration analog/digital conversion assures superior stability, accuracy and noise immunity

Supplied with test leads, AC line cord, operator's manual, without batteries \$295



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INFORMATION RETRIEVAL NUMBER 107

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- 10 or 12 positions, 1 to 6 poles, up to 12 decks.
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- Ideal for instruments, medical electronics, avionics, mobile communications, and many more applications.

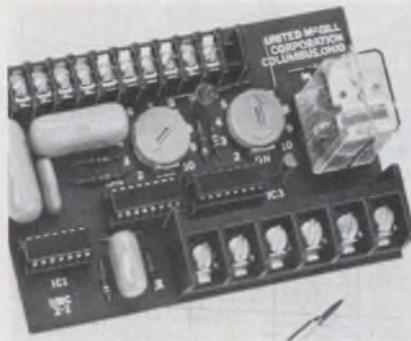
From Grayhill's capability at insert molding technology and value engineering emerges this miniature switch family that provides the performance benefits of enclosed construction at price levels associated with open-wafer. No compromise in materials—even gold plated contacts and diallyl phthalate insulation—in this "best switch for the money" you're likely to find anywhere. For information on these Series 71 switches, consult EEM or write for engineering data.



561 Hillgrove Avenue • LaGrange, Illinois 60525
(312) 354-1040

MODULES & SUBASSEMBLIES

**Long period timer wires
as repeat cycle or delay**



United McGill Corp., P.O. Box 280, 2400 Fairwood Ave., Columbus, Ohio 43216. (614) 443-0192.

The Model UMC-1 Unitimer can be field wired for use either as a time delay relay or as a repeat cycle timer. It can cover time ranges from 0.8 to 40 s or 12 min. to 10 hr. By jumping terminals it is possible to program various timing combinations over a wide range of precisely set on and off intervals (varying from 0.8 s to 10 hr.). Reference dials and a LED are provided for $\pm 2\%$ time setting and checking. The Unitimer has a 5 A spdt (Form C) output that can handle a 120-V-ac resistive load. The device operates on an input of 12 V dc $\pm 10\%$, 70 mA and in temperatures of -10 to $+60$ C.

CIRCLE NO. 327

**Voltage sensor stability
better than 1 mV/°C**

Calex, P.O. Box 555, Alamo, Calif. 94507. (415) 932-3911. \$48; 2 to 3 wk.

The Model 725 ac Volt sensor is a precision ac voltage monitor, in a $1.5 \times 1.5 \times 0.7$ in. epoxy potted module. The unit has a single adjustable 0-to-10-V ac pk-pk trip point range and a sensitivity better than 10 mV. Input impedance is greater than 100 k Ω and trip point stability is better than 1 mV/°C. The dc output is approximately 80% of the power supply voltage at up to 50 mA. A single supply from +12 to +30 V dc with regulation to 5% is required. The unit is also available with the following modifications: reversed output, latching output, or 0 to 100 mV trip point range.

CIRCLE NO. 328

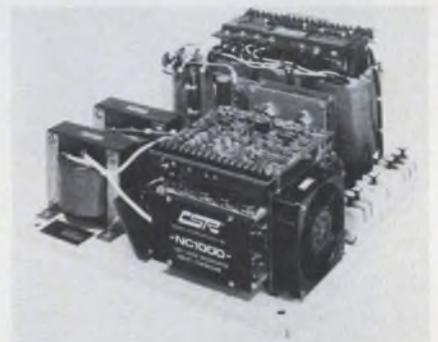
**Hybrid quadrature works
from 0.1 to 10 MHz**

Merrimac, Ind., 41 Fairfield Pl., West Caldwell, N.J. 07006. (201) 228-3890. \$225; 30 to 45 day.

The Model QH-7-4.9 hybrid quadrature is designed for applications in the 0.1-to-10-MHz frequency range. Its 100:1 bandwidth is particularly attractive for such applications as communications systems, single-sideband modulators, antenna-phasing networks, homing systems, and image reject mixers. The coaxial network has four BNC female connectors. Minimum performance specifications include amplitude balance of ± 0.75 dB with a pk-pk amplitude balance of 1.5 dB, nominal coupling of -3 dB, maximum insertion loss typically 1 dB, VSWR of 1.3:1 at 50 Ω impedance, phase quadrature of $90^\circ \pm 3^\circ$, minimum isolation of 20 dB, and average power to 100 mW. The unit's dimensions are $4.25 \times 2.3 \times 1.53$ in.

CIRCLE NO. 329

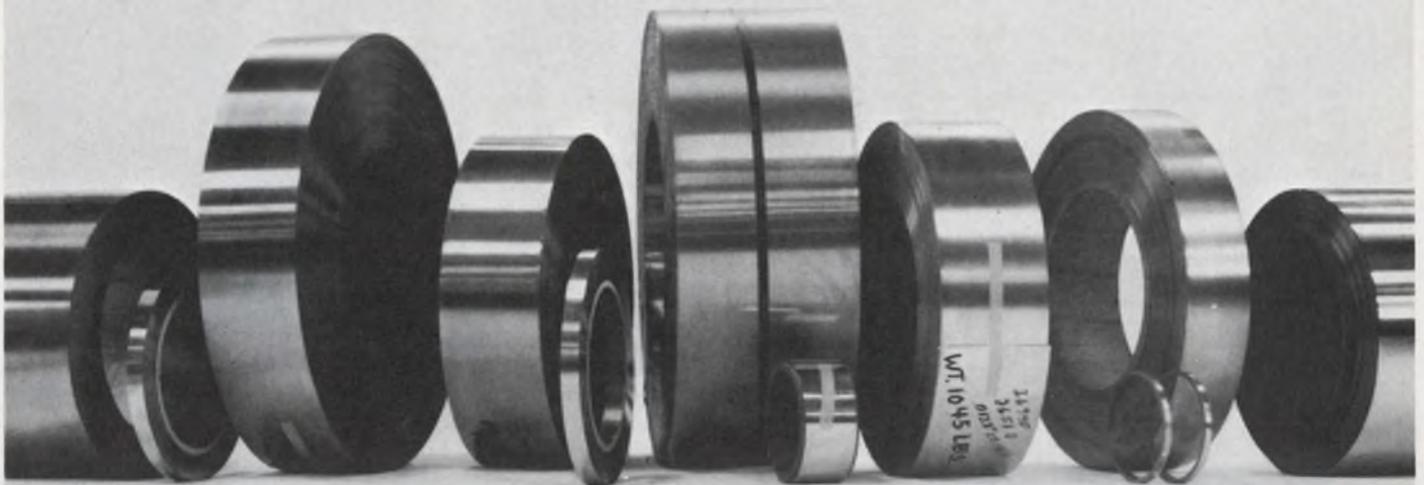
**Modulated motor control
handles 10 to 50 hp**



Control Systems Research, 632 Fort Duquesne Blvd., Pittsburgh, Pa. 15222. (412) 566-1200.

The two-state modulated dc servo controller Model NC1000 can control dc motors in the 10 to 50 hp ranges. Both pulse-width and frequency modulation are incorporated in the controllers. The servo controller has a form factor of 1.01 compared with form factors of 1.4 and above for conventional SCR controllers. The low form factor can increase the efficiency of the dc motor by as much as 60%. The bandwidth of 500 Hz represents an increase by a factor of eight over conventional SCR techniques.

CIRCLE NO. 330



For quick action on metals too special for the big or little mill, come to the mill that's both.

We take on what the big mills can't handle economically or promptly and the small mills haven't the equipment for.

Our specialties are: magnetic alloys, glass-to-metal sealing alloys, expansion alloys, temperature compensating alloys, shielding alloys and electronic alloys. In strip, foil, cut lengths or rod and wire. From several pounds to several tons.

We're geared to give fast service because we're totally integrated.

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MODULES & SUBASSEMBLIES

D/a converter has gain drift of only 10 ppm/°C

Burr-Brown, International Airport Industrial Park, Tucson, Ariz. 85734. (602) 294-1431. \$169 (10 to 24); stock to 2 wk.

The DAC85LD hybrid 12-bit d/a converter has laser trimmed linearity to $\pm 1/2$ LSB. It also has an internal reference and output

amplifier. The gain drift is guaranteed at ± 10 ppm/°C and the offset drift is ± 5 ppm/°C. The DAC-85LD is available in a voltage or current output version. Both are hermetically sealed in a 24-pin DIP case and offer nonlinearities of $\pm 0.012\%$ over an extended temperature range of -25 to $+85$ C. The voltage output model settles to $\pm 0.01\%$ in $5 \mu\text{s}$, and current output model in 300 ns.

CIRCLE NO. 331

Mini charge amplifier requires only 0.5 in.³



Microcom, 1115 Mearns Rd., Warminster, Pa. 18974. (215) 672-6300.

A thick-film, hybrid, high-g charge amplifier, Model MA-95, is designed to interface between piezoelectric transducers and other components in airborne telemetry systems and it takes up only 0.5 in.³ of space. The amplifier is able to respond to changes in charge at the input instead of changes in voltage, and thus produce a high level output. The MA-95 offers optional features including gain and dc offset adjusts and a low-pass filter. The amplifier is encapsulated to withstand high shock levels such as those encountered in gun launch applications.

CIRCLE NO. 332

Test your logic before you test your logic circuit. Logical.



Model 1248
IC Functional Tester

Doesn't it make great good sense to eliminate your malfunctioning digital ICs—TTL, DTL and CMOS—before you wire them in a circuit?

The Model 1248 does precisely that, evaluating each device in from 1 to 5 seconds, for only \$535. And every time you eliminate a disabled IC with your Model 1248, you save yourself from 30 to 60 minutes of expensive logic circuit troubleshooting trying to find the little bugger.

Simply insert the IC in the tester socket, push the

button, and check the digital readout for the proper test code. No comparison with a "good" IC is necessary, and changeover from one device type to another takes less than a minute.

Write for our free full-color brochure on this entirely logical logic tester.

Electro Scientific Industries
13900 N. W. Science Park Drive
Portland, Oregon 97229
Phone: (503) 648-4141



Dc-to-synchro converter current limits to 1 A

Singer Kearfott, 1150 McBride Ave., Little Falls, N.J. 07424. (201) 256-4000.

The C70 4798 402 converter accepts dc signal inputs proportional to the sine and cosine of an angle and provides a three-wire, 400 Hz, 11.8-V synchro stator output. The output (protected against overload or short circuit) can drive multiple loads of torque repeaters or control transformers. The current limiting range of 1 A (nominal) can be extended to 3 A by use of four jumper wires at the connector. The converter has a static accuracy of 0.2 degree and will track input variations of $720^\circ/\text{s}$ within 3° . This unit is packaged as a two-channel unit on a printed circuit card and has dimension of $9.8 \times 7.25 \times 1.12$ in. maximum. It is also available as a single channel unit on special order.

CIRCLE NO. 333

Data-acquisition system provides 32 channels

Data Translation, 109 Concord St., Framingham, Mass. 01701. (617) 879-3595. \$795.

The 32-channel data-acquisition system, Model DT1640/16EX, features a 25 kHz throughput rate, full 12-bit resolution, 0.5-LSB linearity and a system accuracy of 0.03% of full scale range. Two identically sized modules, $3 \times 4.6 \times 0.375$ in., make up of the full 32 channels. A feature of the DT1640/16EX is its immunity to noise when used in close proximity to the minicomputer core memory. Encasement of the modules in steel packaging prevents EMI/RFI noise from causing interference in sensitive analog circuits.

CIRCLE NO. 334

Synchro-to-dc converters accurate to within 0.1%



Astrosystems, 6 Nevada Dr., Lake Success, N.Y. 11040. (516) 328-1600. Stock.

The M2000S series of synchro-to-dc converters can convert selsyn, resolver or synchro signals to proportional dc for input to computers, data loggers and analog or digital displays. The converters are accurate to 0.1%, are packaged in $2.5 \times 2.5 \times 0.8$ in. modules and provide a choice of outputs: 0 to 10 V dc for 0 to 360°, or -5 to +5 V dc for -180 to +180°. The converters may be used with an internal or external reference; they are short-circuit proof and have full transformer isolation at the inputs. Standard models include units that are compatible with 60 Hz synchro and selsyn inputs; 400 Hz synchro or resolver inputs.

CIRCLE NO. 335

Volts My Line?



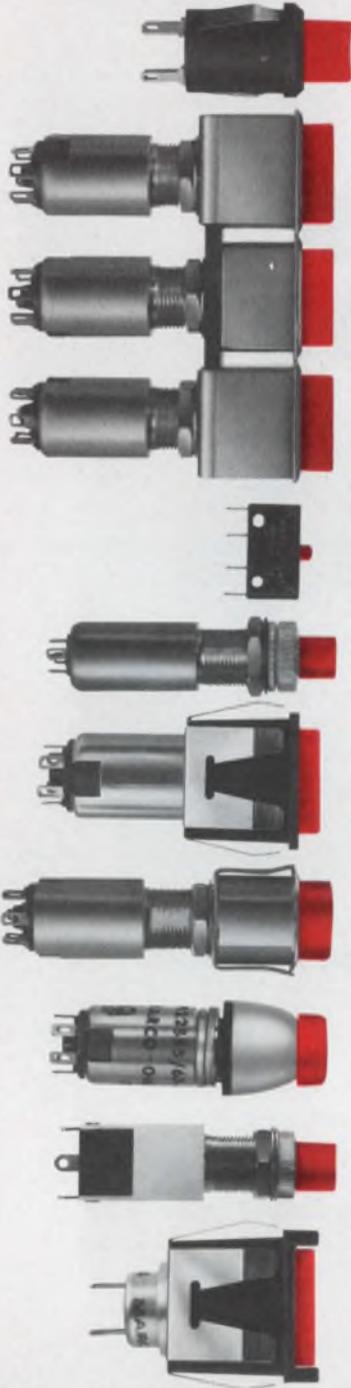
Holding that line-voltage. AC voltage regulation is efficient with the Sorensen ACR Series all solid-state regulators. Problems and damage caused by erratic line voltage are eliminated . . . performance is maximized with ACR features that include MTBF > 25,000 hours . . . remote sensing and remote programming capabilities . . . electronic current limiting . . . adjustable output voltage . . . power outputs to 15kVA—depending on model. Eight models in series. Three-phase systems available. For complete data, contact the Marketing Manager at Sorensen Company, a Unit of Raytheon Company, Manchester, N.H. (603) 668-4500.

Representative Specifications — ACR

- Input Voltage: 95 to 130 Vac
- Total Regulation: 0.4% to 0.6%
- Efficiency: to 95%
- Price Range: \$395 to \$1825

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POWER SUPPLIES

Switches-



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Lighten your decisions
contact . . .

ILLUMINATED PRODUCTS INC.

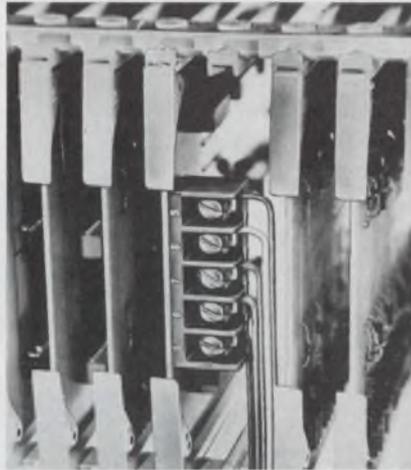
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Tel: (714) 535-6037 TWX 910-591-1185

INFORMATION RETRIEVAL NUMBER 112

COMPONENTS

PC barrier terminals mount at right angles



Reed Devices Inc., 21W183 Hill Ave., Glen Ellyn, Ill. 60137. (312) 858-2050. \$1.17: 10 position (100 up); stock to 4 wks.

A right-angle-mounting, barrier, terminal strip for PC boards affords all the structural strength of regular mounting types without auxiliary support brackets. The 6PCR series, available in two to 24 positions, features 6-32 screws with wire-clamps and specially designed PC terminals on 3/8-in. centers. Labor-saving wire clamps eliminate need for crimp-on terminals as the stripped bare wire is placed under the raised clamp and the screw tightened for a full pressure connection. The strips are shipped with the screws raised.

CIRCLE NO. 336

Bracket converts snap switch to PB assembly

Cherry Electrical Products Corp., 3600 Sunset Ave., Waukegan, Ill. 60085. (312) 689-7600. \$.26 (2000 up).

A new bracket and plunger assembly makes it an easy matter to add a pushbutton to many Cherry snap-action switches. The new unit features a one-piece molded bracket and ferrule, and a choice of gray or red button. The bracket can be supplied as an addition to the regular Cherry Series E33 and E61 switches, as well as the three-wide E61 subminiature modules. The bracket is indicated by adding a "J" to the regular Cherry switch number.

CIRCLE NO. 337

Rocker switches are UL recognized

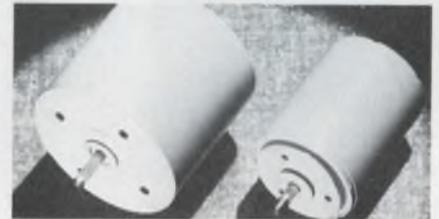


Oak Industries, Inc., Crystal Lake, Ill. 60014. (815) 459-5000. 8 to 10 wks.

A family of UL-recognized rocker switches for the appliance and industrial/commercial market, designated OAK Series 17 rocker switches, are illuminated SPST snap-action switches. All switches have metric dimensions. The switch has a maximum rating of 16 A, 120 V and 8 A, 240 V ac inductive, and it can function in ambient temperatures to 90 C. The body is available in black or white, and the lens in amber, red or clear with other colors on special order. It is available with or without chromium trim.

CIRCLE NO. 338

Miniature motor moves without cogging

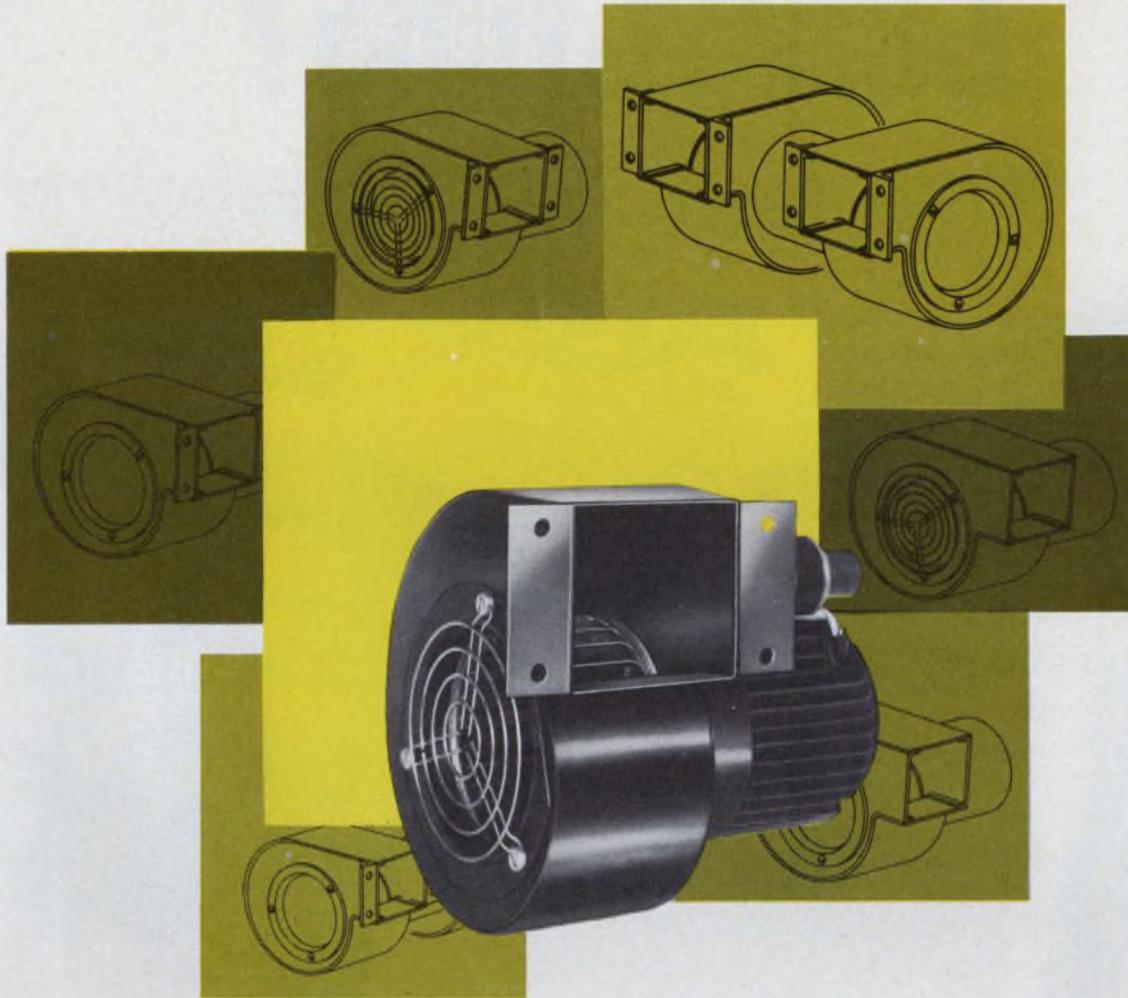


North American Philips Controls Corp., Cheshire Industrial Park, Cheshire, Conn. 06410. (203) 272-0301.

Two miniature, ironless-rotor, reversible, dc motors are ideally suited for reel and capstan drives in digital-cassette recorders, X-Y drives and chart-and-pen drives in chart recorders. They can also be used in magnetic-tape drives and special drives in sophisticated cameras. A 24-V-dc, 3.85-W and a 12-V-dc, 3.7-W model are available. Both offer low cogging, rapid acceleration and low noise levels. Use of the ironless rotor and an oblique-winding design provides a low rotor inertia, high starting torque and a low mechanical time constant. Both motors can achieve efficiencies of over 70%.

CIRCLE NO. 339

Rotron? In commercial blowers?



Ask Eastman Kodak. Or Ampex. Or IBM. Or Control Data. Or Digital Equipment.
Or Honeywell. Or Univac. Or Mohawk Data Sciences.

These are a few of the dozens of companies that are major customers for Rotron's Vanguard™—the new commercial centrifugal blower that provides a balanced combination of quality and economy. In simplex and duplex models, flows to over 600 cfm, pressures to 2" H₂O. And that can be quickly and easily customized to meet highly specific customer needs.

Many of these same companies, of course, have long known Rotron for its leadership in fan products. For the completeness of its application engineering and distributor services. And for the fact that these Rotron "extras" cost no more.

Now they, and you, know this. Rotron® is fast becoming a name in blowers, too.



ROTRON INC.

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Pacific Div., Burbank, Cal. 91506. 213 • 849-7871 • Rotron B.V., Breda, Netherlands. Tel.: 793111, Telex: 844-54074

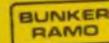
16340

INFORMATION RETRIEVAL NUMBER 113

LOOKING FOR CONNECTORS?



VIKING INDUSTRIES, INC.
STANDARD PRODUCTS DIVISION



AMPHENOL



Electrical
Components
Division

AMP
INCORPORATED

SWITCHCRAFT



ROBINSON
NUGENT INC.



Jermyn

Scotchflex®



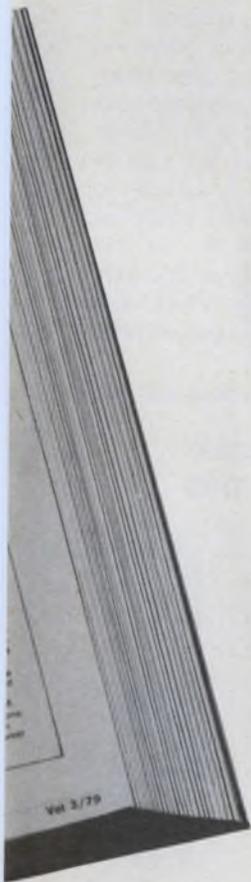
CAMBION

DALE



AUGAT

LOOK FIRST IN Electronic Design's GOLD BOOK



IT'S ALL THERE For example for CONNECTOR PRODUCTS you'll find source information listed for the following categories in the PRODUCT DIRECTORY (each listing gives *complete* manufacturer's name, street address, city, state, zip and phone for convenience in contacting suppliers):

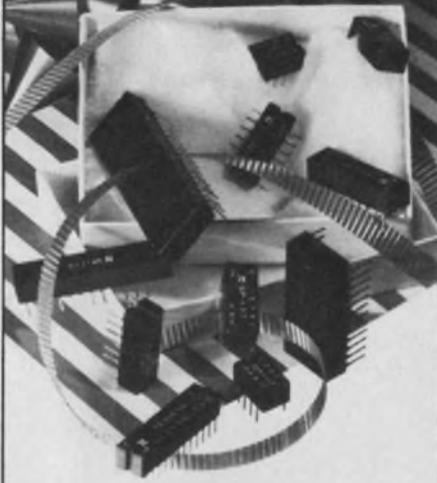
CONNECTOR ADAPTERS • CONNECTOR BOXES AND STRIPS • CONNECTOR CAPS • ANTENNA CONNECTORS • AUDIO CONNECTORS • BATTERY CONNECTORS • CIRCULAR CONNECTORS • COAXIAL CABLE CONNECTORS • CORD CONNECTORS • FILTERED CONNECTORS • HIGH VOLTAGE CONNECTORS • INTERLOCK CONNECTORS • PRINTED CIRCUIT CONNECTORS • QUICK DISCONNECT & PUSH-ON CONNECTORS • RACK & PANEL CONNECTORS • RF CONNECTORS • SOLDERLESS CONNECTORS • STRIPLINE CONNECTORS • UMBILICAL CONNECTORS

YOU'LL FIND 216 CATALOG PAGES ON WIRE & CABLE AND CONNECTOR PRODUCTS

Volume 3 of *Electronic Design's* GOLD BOOK leads off with catalog pages on wire and cable and is followed immediately by connector products and terminal boards. Forty-three manufacturers are represented in these two sections. Amphenol has 24 pages packed with detailed specifications; AMP Incorporated emphasizes its flexible flat cable products, miniature and pin socket connectors; Augat has 10 pages; Robinson Nugent, Inc. offers 8 pages, and so on. Check these pages first before you contact a supplier. You may find the exact item you need—and that goes for each one of the 52 product categories in *Electronic Design's* GOLD BOOK. There are 2,820 catalog pages in all right at your fingertips for immediate reference. And TWICE the number of directory pages available from any other industry source.

**WHEN YOU'RE LOOKING FOR MANUFACTURERS,
DISTRIBUTORS, OR PRODUCT INFORMATION...
ELECTRONIC DESIGN'S GOLD BOOK
IS THE PLACE TO LOOK**

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dip DELAY LINES

Stop looking for DIP packaged electromagnetic delay lines . . . you've found 'em! Mini dip's, standard dip's, low profile dip's and double dip's . . . all with unmatched performance.

| Leads | Length | Width | Height |
|-------|--------|-------|--------|
| 8* | .510 | .300 | .225 |
| 14* | .750 | .300 | .225 |
| 16* | .850 | .300 | .225 |
| 20* | 1.050 | .300 | .225 |
| 24* | 1.250 | .300 | .225 |
| 28* | 1.450 | .300 | .225 |
| 14* | .750 | .300 | .175 |
| 16* | .850 | .300 | .175 |
| 24 | 1.300 | .600 | .235 |
| 34 | 1.975 | .600 | .235 |

*Also available in flat pack configurations.

SPECIFICATION HIGHLIGHTS

- Tap capabilities from 500 ps to 25 ns
- Delays from 5 ns to 1000 ns
- Rise time ratios from 5:1 to 30:1
- Std. impedances of 50, 75, 100 ohms
- Frequency capabilities to 300 MHz
- Temp. -55 to +125°C (TC 70 PPM/°C)
- Manufactured to MIL-D-23859; environmental to MIL-STD-202

Phone, wire or write for more information today!

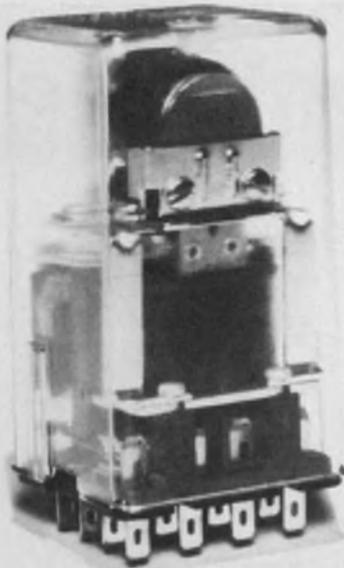


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INDUSTRIES INC.**

22119 S. Vermont Ave., Torrance, Ca. 90502
(213) 325-7440 • TWX 910-349-7512

COMPONENTS

Latching relay handles 10 A with 4PDT contacts



Struthers-Dunn Inc., Lambs Rd.,
Pitman, N.J. 08071. (609) 589-
7500.

Mechanical latching and electrical reset are featured in Model 455 relays for 10-A service. The relay has 4PDT load-handling contacts capable of controlling 1/3-hp motor loads or 10-A resistive loads at 120 V ac. After the operate coil is energized, the contacts remain transferred, even if power fails, until reset. For test or maintenance, optional manual actuators are available for both operate and reset. Voltages for the reset and operate coils are 6 to 240 V ac or 6 to 125 V dc. Contacts have a minimum life expectancy of 300,000 operations at rated loads.

CIRCLE NO. 340

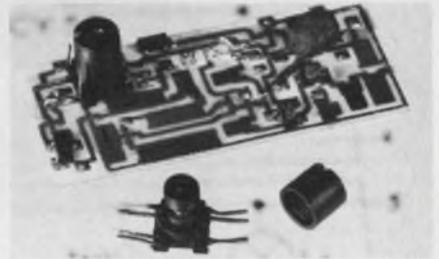
DIP dry-reed relay handles power voltages

Magnecraft Electric Co., 5575 N.
Lynch Ave., Chicago, Ill. 60630.
(312) 282-5500.

Magnecraft's new eight or 14-pin Class 171VDIP dry-reed DIP relay can switch 120 V rms or 200 V dc. This allows both triac triggering and compatibility with DTL and TTL combined with total isolation of the integrated circuit. The relay is available in 1-Form A contact arrangement with coil voltages of 5, 6, 12, or 24 V dc. All versions provide for a choice of internal clamping diodes.

CIRCLE NO. 341

Miniature cores provide Qs to 200 at 200 kHz

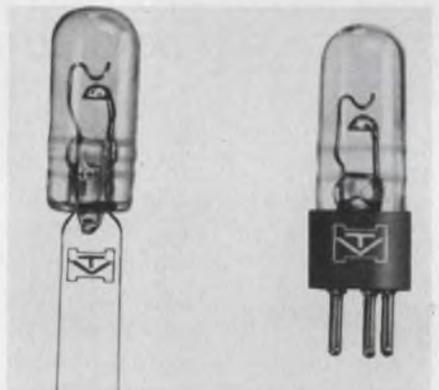


Siemens Corp., 186 Wood Ave. S.,
Iselin, N.J. 08830. (201) 494-
1000. \$0.90 (1000 up).

Conventional discrete inductors are often required for miniature electronic circuits. Siemens' new adjustable miniature core can be used for resonant circuits, chokes and transformers. It consists of a yarn roller core with a screw-on cup core. When the cup core is attached, the completely enclosed unit can be dip soldered on a PC board. Values of Q as high as 200 can be achieved. The core is available in three different materials covering frequency range from 0.1 to 200 MHz. A Q factor of 175 at 200 kHz and 50 MHz with 0.23-mH and 0.15- μ H inductances, respectively, is typical.

CIRCLE NO. 342

Ultraviolet detector responds to flames



Hamamatsu Corp., 120 Wood Ave.,
Middlesex, N.J. 08846. (201) 469-
6640.

The R334 is a head-on detector that responds only to ultraviolet radiation between 1600 and 2900 Å with maximum sensitivity at 2100 Å. It is ideal for detecting flames in a fire alarm or in a fail-safe, flame-control monitor. The detector can recognize if the pilot flame is out or if the fuel supply is incorrect.

CIRCLE NO. 343



“A good general purpose signal generator should handle most of your bench needs and be cheap enough to fit every budget. That’s why we designed the HP 8654A”

Frequency range: 10-520 MHz
Calibrated
output level: +10 to -130 dBm, leveled
Modulation: AM & FM,
internal & external
Stability: 20 ppm/5min
Price: \$1700. Domestic USA price only.

But you'll want to evaluate the 8654A for yourself. A call to your local HP field engineer will put one in your hands. Or for more information just write.

HEWLETT  PACKARD

Sales and service from 172 offices in 65 countries.
1501 Page Mill Road, Palo Alto, California 94304

Mercury relay sells for less than \$5

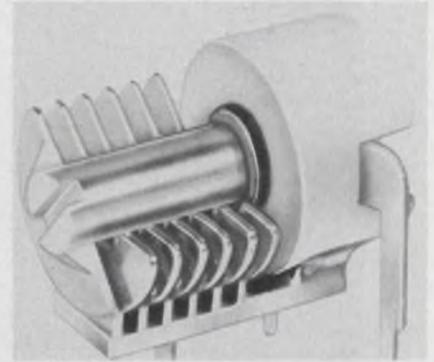
Durakool, Inc., 1010 N. Main St., Elkhart, Ind. 46514. (219) 264-1116. See text.

Durakool's 10-A model of its new series of small mercury-to-mercury displacement relays costs less than \$5, and the 15 and 20-A

models cost only slightly more. The new relays offer a range of coil voltages from 24 to 600 V ac or dc. The units are ultra compact. They measure only 1.5 × 1.75 × 2.0 in. The stainless-steel body minimizes overload damage because it permits use of hydrogen under high pressure to reduce arcing. And fully enclosed contacts and hermetic seals ensure against explosions and fires.

CIRCLE NO. 344

Variable air capacitor mounts on PC board



E.F. Johnson Co., Waseca, Minn. 56093. (607) 835-2050.

A subminiature Type T variable air capacitor has a horizontal PC mount that gives the tuning capacitor more stability and strength. The rotors and stators are precision machined from solid-brass extrusions for stability and uniformity. The capacitor is low in cost (not given) and about one-third the volume of a U capacitor. Metal parts are silver plated and iridized for easy soldering. Ceramic is 95% alumina, Grade L-624 or better. Temperature coefficient is +30 ± 15 ppm/°C. High torque holds rotor securely under vibration. Peak voltage is 250 V.

CIRCLE NO. 345

No need to look further for a reliable R-F power source. MCL has one.



And here's what we built into it:

- solid state circuitry
- short and open circuit protection
- frequency stability ±0.1 db
- external pulse or AM modulation
- internal square wave modulation
- low tube cost/operating hour
- qualified to MIL-STD-461 and 810

It has six different plug-in heads

| Model | Freq. (MHz) | Pwr. (MIN) |
|-------|-------------|------------|
| 6047 | 10-50 | 65 |
| 6048 | 50-200 | 65 |
| 6049 | 200-500 | 65 |
| 6050 | 500-1000 | 65 |
| 6051 | 1000-2000 | 40 |
| 6052 | 2000-2500 | 25 |

You have now ended your search for a stable, reliable 65 watt oscillator. Just call or write for detailed engineering data. Or ask for a demonstration

MCL, INC.,
10 North Beach Avenue,
LaGrange, Illinois, 60525.
(312) 354-4350



Now on GSA contract GSOOS-27086 See us in EEM—Vol. 1 pp. 284-291

PB-switch kit allows custom design assembly

Ledex Inc., 123 Webster St., Dayton, Ohio 45401. (513) 224-9891. \$25; stock.

To shorten prototype time, a new pushbutton-switch designer's kit containing 135 parts is offered by Ledex. The user can custom design up to a six-station Series 1400 switch assembly with interrelated switching logic. The designer can build momentary and push-push switches with interlock, reciprocal release, master release or any combination of these functions. The kit includes step-by-step assembly instructions and identification drawings for ten different release functions. Also included are assortments of pushbutton caps, lens, mounting rails, blocks, release rods and six switch modules. The kit can be replenished. An enclosed order form gives a complete description, part number and price for each component in the kit.

CIRCLE NO. 346

Tall and narrow oven conserves floor space

The Grieve Corp., 510 Hart Rd., Round Lake, Ill. 60073. (312) 546-8225.

The high and narrow type P-402 electric oven can handle long pieces and conserves floor space. The unit's inside dimensions are only 30 × 38 in., but it is 108-in. high. Channels for up to eight shelves are spaced on 13-1/2-in. centers, which permits curing several shelves of smaller parts in addition to long pieces. A maximum temperature of 650 F is achieved from a 30-kW heat input that is circulated upward at 2450 cfm. Controls include an indicating proportioning temperature controller, a manually resettable excess temperature controller, and a 12-h timer for automatically shutting off the oven after a specified period of operation.

CIRCLE NO. 347

Diamond saw dices clean edged alumina chips



Aremco Products, Inc., P.O. Box 429, Ossining, N.Y. 10562. (914) 762-0685. See text.

A diamond dicing-saw system cuts alumina substrates for use in thick and thin-film circuits and provides an improved edge quality, according to the manufacturer. Because of delays in obtaining alumina substrates to exact size and tolerance, hybrid-circuit makers have been shifting to dicing their own substrates. Thus, the user can stock 2 × 2 or 1 × 1-in. substrates and dice them to size. But laser dicing machines cost about \$40,000 and ordinary diamond mechanical scribing, though inexpensive, produces poor edges. However, Aremco's Accu-Cut 5000 diamond dicing saw system costs less than \$5000.

CIRCLE NO. 348

GET ON BOARD...

with TANTALEX[®] Low-Cost Solid Tantalum Capacitors

Sprague Gives You a Choice for Flexibility in Your Printed Board Designs.



DIPPED . . . SPRAGUE TYPE 196D

Small size economical capacitors that utilize high-quality tantalum pellet construction. Conformal epoxy resin coating is highly resistant to moisture and mechanical damage. This capacitor has found wide usage in consumer and commercial electronic equipment. Operating temperature range, -55C to +85C. Available in all popular 10% decade values from 0.1 μF to 330 μF. Voltage range, 4 to 50VDC. Standard lead spacing, 0.125" and 0.250". For complete data, write for Engineering Bulletin 3545B.



MOLDED . . . SPRAGUE TYPE 198D

Economically priced, molded-case Econoline™ capacitors. Standard lead spacing, 0.100", 0.200", and 0.250". Tough, flame-retardant, crack resistant case has flattened section and polarity indicator for easy-to-read marking and error-free insertion. Fixed external dimensions allow increased productivity during assembly of PC boards. Designed for severe vibration and shock environment, where lead support alone is not adequate. Operating temperature range, -55C to +85C. Capacitance values from 0.1 to 100 μF. Voltage range, 4 to 50VDC. For complete data, write for Engineering Bulletin 3546.

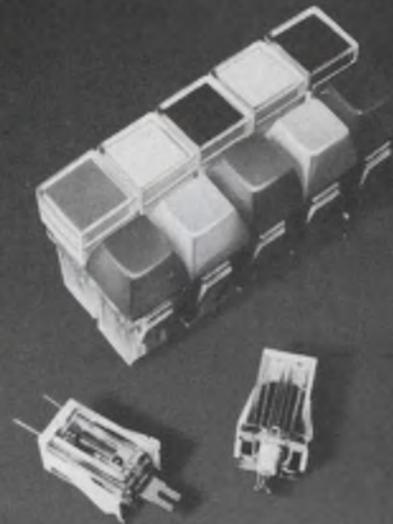
Call your nearest Sprague district office or sales representative, or write for the bulletins mentioned above to Sprague Electric Company, 347 Marshall Street, North Adams, Mass. 01247.

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Contributing an extra feature to the automated systems such as computer peripheral equipment, measuring instruments, vending machines, control appliances, etc.

Built-in are a reed switch and a bi-pin lamp. Switching is perfect and the lamp shows automatically whether the circuit is active. Installation can be completed at once by a single push-in action.

PARTICULARS

Operating force: 70g, full stroke

Stroke: 4mm, full

Colors: Snow white, bright green, yellow, true blue, and fire red

Rated voltage: 5, 12, and 28V

Average service life: 50,000 hrs. or better

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Tokyo 142, Japan Phone: 785-1111
Telex: 2466301 SMK TOK J
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PACKAGING & MATERIALS

Lab ware is inert to molten metals, halides



Union Carbide Corp., 270 Park Ave., New York, N.Y. 10017. (212) 551-5114.

Semiconductor materials are temperamental when it comes to impurities. They must be processed under extremely sterile conditions. High-purity lab ware is an absolute necessity. Laboratory crucibles, evaporating dishes, tubes and metallizing boats are among the standard items available. Custom shapes can be fabricated to user specifications. They are made from Boralloy pyrolytic boron nitride, which is inert to most molten metals and halide salts. The material has the high thermal-shock resistance needed for repeated cycling to temperatures as high as 2000 C. Vessels made from it can't contaminate critical material that is processed in them, according to Union Carbide.

CIRCLE NO. 349

Spreader pad converts TO-5 relay to DIP

Teledyne Relays, 3155 W. El Segundo Blvd., Hawthorne, Calif. 92050. (213) 973-4545.

Spreader pads for TO-5-cased relays provide a standard DIP pin-out of 0.3-by-0.1-in. spacing. TO-5 relays can be inserted and soldered to standard-grid PC boards or plugged into 14 or 16-pin DIP sockets. Teledyne relay users can order relays with the spreader pads mounted.

CIRCLE NO. 350

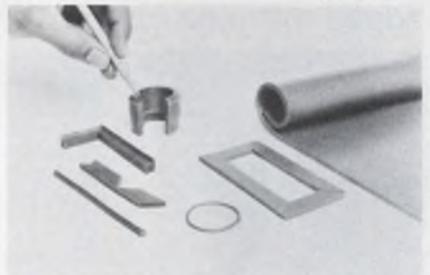
Thick-film kit provides trial quantities

Cermet Div. of Bala Electronics Corp., 14 Fayette St., Conshohocken, Pa. 19428. (215) 828-4650. \$195; stock.

Trial-quantity evaluation kits of thick-film pastes for screen-printable electro-optics contain four conductor materials and two dielectric materials. The materials are high and low-temperature-firing nickel conductors, copper and silver conductors, an opaque dielectric and a clear dielectric. These pastes are designed for applications that include the production of LEDs, liquid crystals and plasma displays, channel-multiplexers, fiber optics and ultraviolet transmitting faceplates. The pastes fire in a nitrogen atmosphere and thus eliminate the possibility of explosion associated with hydrogen processing. Each jar contains 1 oz of thick-film material.

CIRCLE NO. 351

Conductive elastomer provides EMI seals

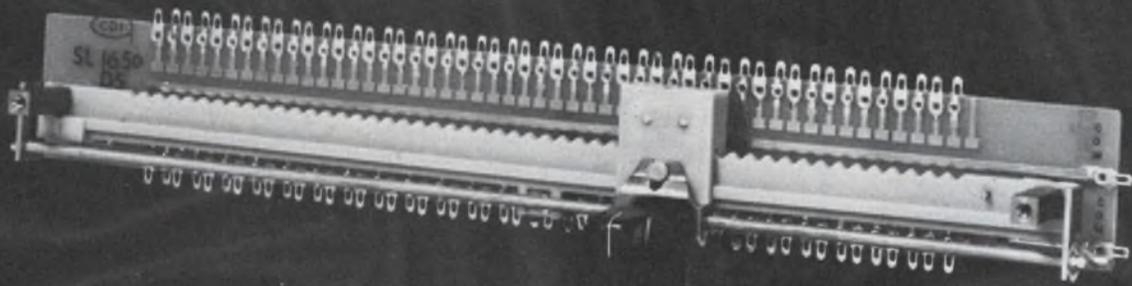


Technical Wire Products Inc., 129 Dermody St., Cranford, N.J. 07016. (201) 272-5500. \$15 to \$30 per sheet (15 up).

Conmax, an electrically conductive elastomer, is particularly suited for EMI shielding and environmental sealing applications. It is a high-performance, low-compression-set silicone elastomer (ZZ-R-765, Class III), which contains a uniform dispersion of randomly oriented nickel fibers to impart electrical conductivity. The material is compatible with most metals and alloys. Standard and custom sheets from 0.020 to 0.125-in. thick and die-cut or stamped gaskets in strips or molded form are available. The material has an operating-temperature range of -65 to 450 F. Standard sheets are 12 x 12 x .020 to 0.060-in. thick.

CIRCLE NO. 352

CDI CREATES THE BETTER LINEAR SWITCH



CHOOSE ANY NUMBER OF POSITIONS UP TO 100 (UNIQUE)

Series SL Linear Slide Switch (Patented)

APPLICATIONS: CONTROL FUNCTIONS REQUIRING RAPID POSITION SELECTING

- CATV channel selecting
- High & Low limit switches
- Process equipment controls
- Computers & peripherals
- NC applications
- Many others

OTHER REMARKABLE FEATURES:

Mounts EITHER left/right OR up/down (versatile) • Single or multiple position selectors • Solder lugs or dip solder termination for p.c. use • 1 or 2 Poles, or coded outputs available.

ELECTRICAL CHARACTERISTICS:

200 megohms min. insulation • 1000 volts min. dielectric strength • 2 amps @ 115 VAC current carrying capability • 125 ma @ 115 VAC current breaking capability.



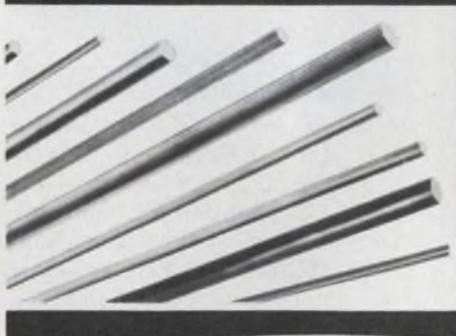
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PRECISION PRODUCTS DIVISION

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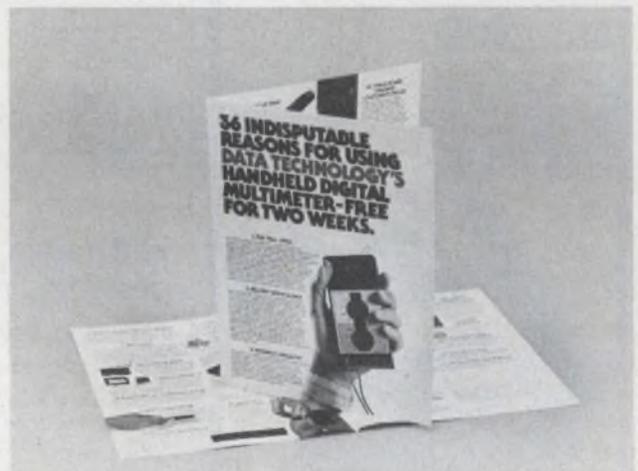
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ELECTRONIC DESIGN 20, September 27, 1974



IF YOU MISSED IT LAST ISSUE, GET IT NOW.

If you missed the Data Technology Model 21 insert from the last issue, or if it was ripped out before you got the magazine, just circle the reader service number below, and we'll rush you a copy.

The Model 21: It's worth reading about. A handheld portable digital multimeter that measures AC volts, and DC volts, ohms and capacitance. It's the only handheld DMM that measures capacitance.



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...in test systems where integrity is essential, specify Cunningham



Crossbar switches

FROM TEN TO THREE HUNDRED 6 WIRE CROSSPOINTS. EXTREMELY LOW PATH RESISTANCE. LOW NOISE AND MINIMAL THERMAL EMF. MINIMUM OF 20 MILLION OPERATIONS/CROSSPOINT LIFE EXPECTANCY. HANDLES LOW LEVEL AND BROADBAND SIGNALS.



reed matrix

STANDARD 5X10 (50 CROSSPOINT) MATRIX FOR GROUP TO GROUP SWITCHING. INTERMEDIATE FREQUENCY (IF) HANDLES DC TO 4 MHz SIGNALS. HIGH FREQUENCY (HF) CAPABILITY OF 60 MHz WITH SHIELDED SIGNAL PATHS.



matrix controller

INTERFACES A SWITCH MATRIX TO ANY 16 BIT ADDRESS DEVICE. FACILITATES A WIDE VARIETY OF SWITCH MATRIX APPLICATIONS BY PROGRAMMABLE REMOTE AND MANUAL CONTROL. ESPECIALLY USEFUL IN AUTOMATIC TESTING OR TELEMETRY SWITCHING.



scanners

FOR SEQUENTIAL OR RANDOM CONNECTION OF INSTRUMENT LEVEL SIGNALS. REED SCANNERS OFFER HIGH SPEED SWITCHING FOR DATA LEVELS. CROSSBAR SCANNERS BEST FOR LOW THERMAL OR LOW NOISE DATA ACQUISITION SYSTEMS. BOTH FEATURE LOCAL AND REMOTE CONTROL.

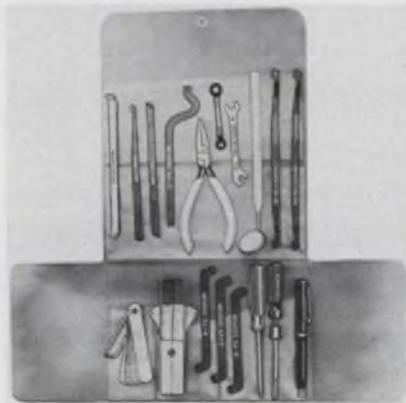
applications assistance

FOR ADDITIONAL DATA OR APPLICATIONS ASSISTANCE, CALL THE CUNNINGHAM CORPORATION, HONEYE FALLS, NEW YORK: (716) 624-2000.

Cunningham Corporation
SUBSIDIARY OF GLEASON WORKS

PACKAGING & MATERIALS

Special tool kit repairs relays



P. K. Neuses, Inc., Box 100, Arlington Heights, Ill. 60006. (312) 253-6555.

There has always been a need for a set of tools specifically designed to service relays. The TK-18 kit is not just a lot of makeshift gadgets to help fill up a tool box, but rather a set of tools exactly suited for any job required in the servicing, aligning, checking and maintenance of relays. The kit comes complete with a vinyl case that includes contact burnisher with extra blades, spring tension gauge, wrenches, thickness gauge set, inspection mirror—everything needed to do the job fast. Included with kit is a new eight-page instruction booklet on "How to Service Relays."

CIRCLE NO. 353

Flat cable zip-on jacket provides rfi shielding

The Zippertubing Co., 1300 S. Broadway, Los Angeles, Calif. 90061. (213) 321-3901.

Designed to fit and protect flat cable, a new zip-on, shielded, flexible jacket, ZFT, is especially suited for use in communications and computer data terminals. The lightweight jacket shields ribbon or flat cable against rfi and electrostatic interference. Standard jacket material is vinyl-coated nylon mesh with 3-mil aluminum-foil shielding. Where extra conductivity is desired, a silver coat is available. A tinned-copper braid runs the full length to provide a positive ground with a solderable termination.

CIRCLE NO. 354

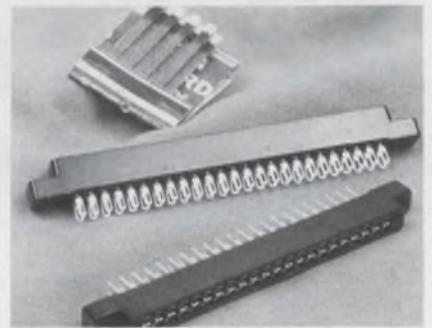
Fabricate conductive rubber items normally

Mobay Chemical Co., Pittsburgh, Pa. 15205. (412) 923-2700.

Semiconductive silicone-rubber material, Baysilone HV 8/700, can be used in the manufacture of conductive flexible-rubber items with most common rubber fabrication techniques, such as molding and extruding. Volume resistivity of the rubber is usually kept within 30 to 50 Ω cm to reduce the effect of mechanical processing on the rubber's resistivity. However, volume resistivity can be increased as high as 10^{16} Ω cm by blending with other nonconductive Baysilone rubber materials. The rubber has a useful temperature range from -50 to 200 C and is available in catalyzed or uncatalyzed form.

CIRCLE NO. 355

PC connector line fits commercial applications



Amphenol Industrial Div., 1830 S. 54th Ave., Chicago, Ill. 60650. (312) 652-1220.

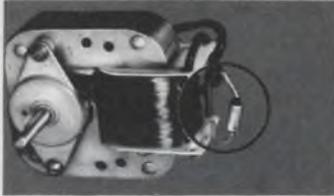
The 117-deviation PC connector line is an expansion of Amphenol's family of 225 Series PC connectors. The line includes 10 popular contact positions—from 6 to 43 positions—to suit most commercial applications. Solder terminations provide for flush installation on PC boards 0.054-to-0.071-in. thick. There are four different tail styles. An eyelet solder tail accepts up to three 20 AWG wires and the other tail styles are straight for dip or flow soldering. All dip-solder terminals are designed for 0.043-in. min diameter holes. The connectors have an operating voltage range from 600 at sea level to 150 V ac rms at 70,000 ft. Current rating is 5 A. Contact resistance produces 30 mV at rated current for gold over copper.

CIRCLE NO. 356

1/32 of an ounce of prevention for your electronic circuits.

MICROTEMP SAFETY THERMAL CUTOFFS*

With a MICROTEMP installed on the winding of your motor or transformer, you'll get positive fail-safe protection without adding



any appreciable weight or size. Facilitating transformer and motor compliance with many UL specifications, especially UL506, UL544, and UL547, MICROTEMP will sense excessive temperature and interrupt the power circuit



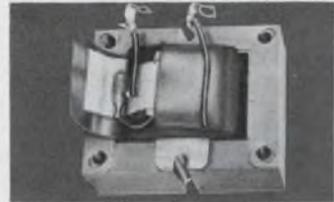
ACTUAL SIZE

should a fault condition develop. An alternate means of protecting the transformer can be provided by a MICROTEMP which has a small heating element fitted to



its case. This heating element is connected in series with the secondary coil winding and its load, transferring heat to the MICROTEMP with the flow of current. Should the current flowing through the element exceed normal operating levels, the resulting heat will cause the

MICROTEMP to open, thus also opening the primary or secondary coil winding to which it is connected. Reliable, inexpensive and accurate to $\pm 3^\circ\text{F}$, MICROTEMPS are available in ratings to fit your application. Add our little prevention to your product now!



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Dayton, Ohio 45419
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INFORMATION RETRIEVAL NUMBER 124

IT'S YOUR MOVE TO BETTER MATRIX BOARDS

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| .125 CENTERS | RELIABILITY • ECONOMY AND STANDARDIZATION | | UP TO 8 DECK SWITCHING |
| .250 CENTERS | .100 CENTERS | BUSSED CONTACT | DUAL CONTACT STRIPS |
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A DIVISION OF LVC INDUSTRIES INC.

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ELECTRONIC DESIGN 20, September 27, 1974

If you think there are no other ways to cut your **POWER SUPPLY** costs, get proven performers and realistic delivery... you didn't figure on

QUAD-MOD
Versatile multiple output
OEM Power Supplies

from **\$135.00**

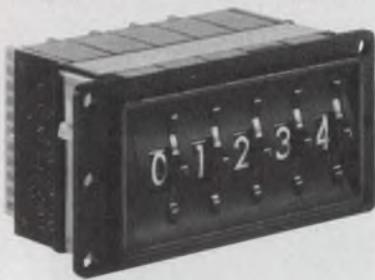


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Price is the Object.



**(25¢ per
switch function*)**

SERIES 29000 ECONOMY MINISWITCH

...has many of the qualities of the more expensive switches with the added feature of being dust and moisture sealed.

A unique assembly strap makes it possible to build assemblies consisting of end mounting brackets and switch modules without using tools.

*List prices start at \$3.00 per switch module or only 25¢ per switch function for a sealed switch with 12 dial positions. That's less than the cost of most toggle, pushbutton, lever, slide, or rotary switches.

YES... "PRICE IS THE OBJECT", but when you buy Series 29000 Economy MINISWITCHES you get performance and quality too. Compare... Evaluate... We think you will agree. Send for details today.

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See pages 1148 and 1149, Vol. 2, of the 1973-74 EEM Directory for more DIGITRAN products.

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INFORMATION RETRIEVAL NUMBER 127

application notes

Spring pins

"70 Ways to Cut Costs and Simplify Assembly Operations," a 32-page booklet with 70 simplified "how to" drawings, illustrates time and cost-saving applications of spring pins. Amerace Corp., Esna Div., Union, N.J.

CIRCLE NO. 357

Metal-film resistors

A paper, "The Reliability of Laser-Trimmed Screen-Printed Metal-Film Resistors," includes photos, diagrams and tables to support the text. Sprague, North Adams, Mass.

CIRCLE NO. 358

Valve-noise booklet

How to calculate valve noise is covered in a booklet. It explains the causes and remedies for excessive noise related to control valve use. Honeywell Process Control, Fort Washington, Pa.

CIRCLE NO. 359

Limiters

"The Application of Limiters to Communications Equipment" discusses limiters for receiver protection, electrical parameters and design constraints. A comparison of the performance of a communications limiter with that of a typical EW limiter is included. Five graphs provide illustrative support. American Electronics Laboratories, Lansdale, Pa.

CIRCLE NO. 360

Inductors, transformers

"Transformer and Inductor Design" is a source book for design and core selection information on toroidal and laminated power inductors and transformers. Representative chapters of the 36-page manual include metric system conversion data, powder core selection, wound toroid O.D. calculation, permeability and air gap and specifications with nomographs. A nominal charge of \$1 per copy is requested. Magnetics, P.O. Box 391, Butler, Pa. 16001.

INQUIRE DIRECT

vendors report

Annual and interim reports can provide much more than financial-position information. They often include the first public disclosure of new products, new techniques and new directions of our vendors and customers. Further, they often contain superb analyses of segments of industry that a company serves.

Selected companies with recent reports are listed here with their main electronic products or services. For a copy, circle the indicated number.

Solid State Scientific. Semiconductors, rf transistors, CMOS and electronic digital watch modules.

CIRCLE NO. 361

P. R. Mallory & Co. Batteries, electrolytic and ceramic disc capacitors, alumina substrates, hybrid circuitry, piezoelectric devices, metals and materials, controls, timers and motors.

CIRCLE NO. 362

ACDC Electronics. Power supplies.

CIRCLE NO. 363

Cognitronics. Optical character-recognition equipment and precision machine parts and subassemblies.

CIRCLE NO. 364

General Radio. Network analyzers, frequency synthesizers, sound-level meters, logic-circuit analyzers, electronic and medical instrumentation and electronic stroboscopes.

CIRCLE NO. 365

Cryogenic Technology. Environmental protection systems, cryogenic systems, cryocoolers and cryogenic vacuum pumps.

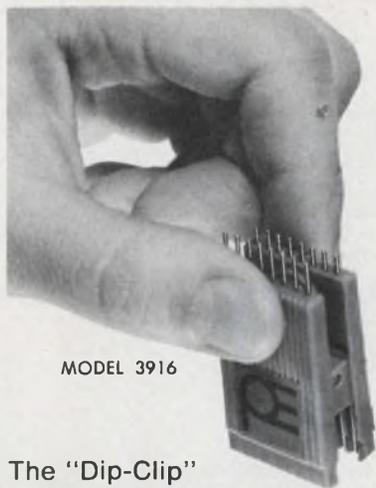
CIRCLE NO. 366

Rapidata. Remote access computer service.

CIRCLE NO. 367

Matsushita Electric. Consumer equipment and components, communication and measuring equipment, lighting equipment, tubes and semiconductors, industrial equipment and batteries.

CIRCLE NO. 368

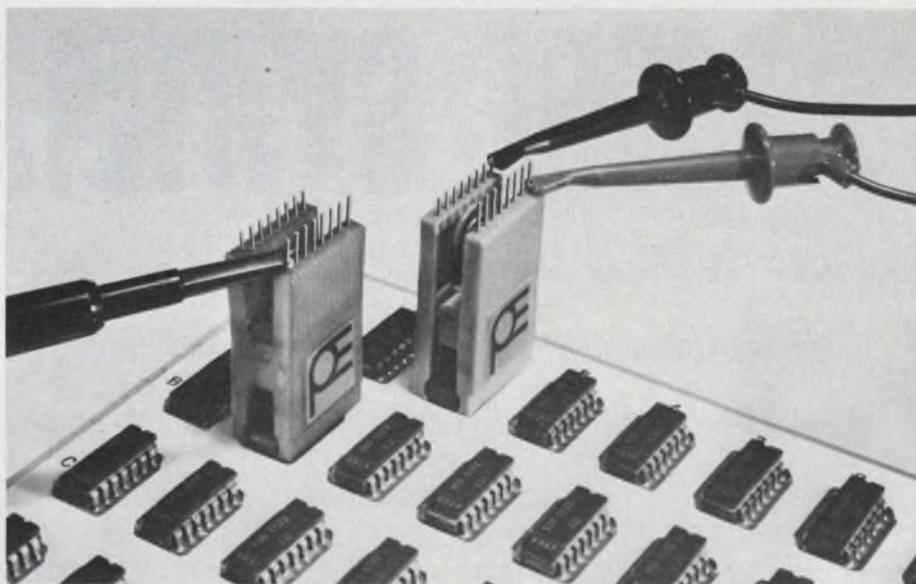


MODEL 3916

The "Dip-Clip" is specially designed to allow the attachment of test probes to 14 or 16 lead DIPs. The unique patented design greatly reduces the possibility of accidental shorting while testing live circuits. Numerous test probes may be quickly connected for hands-free testing.

dip clip

T.M.

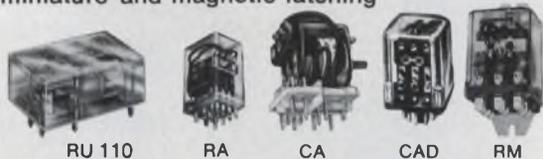


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ELECTRONIC DESIGN 20, September 27, 1974

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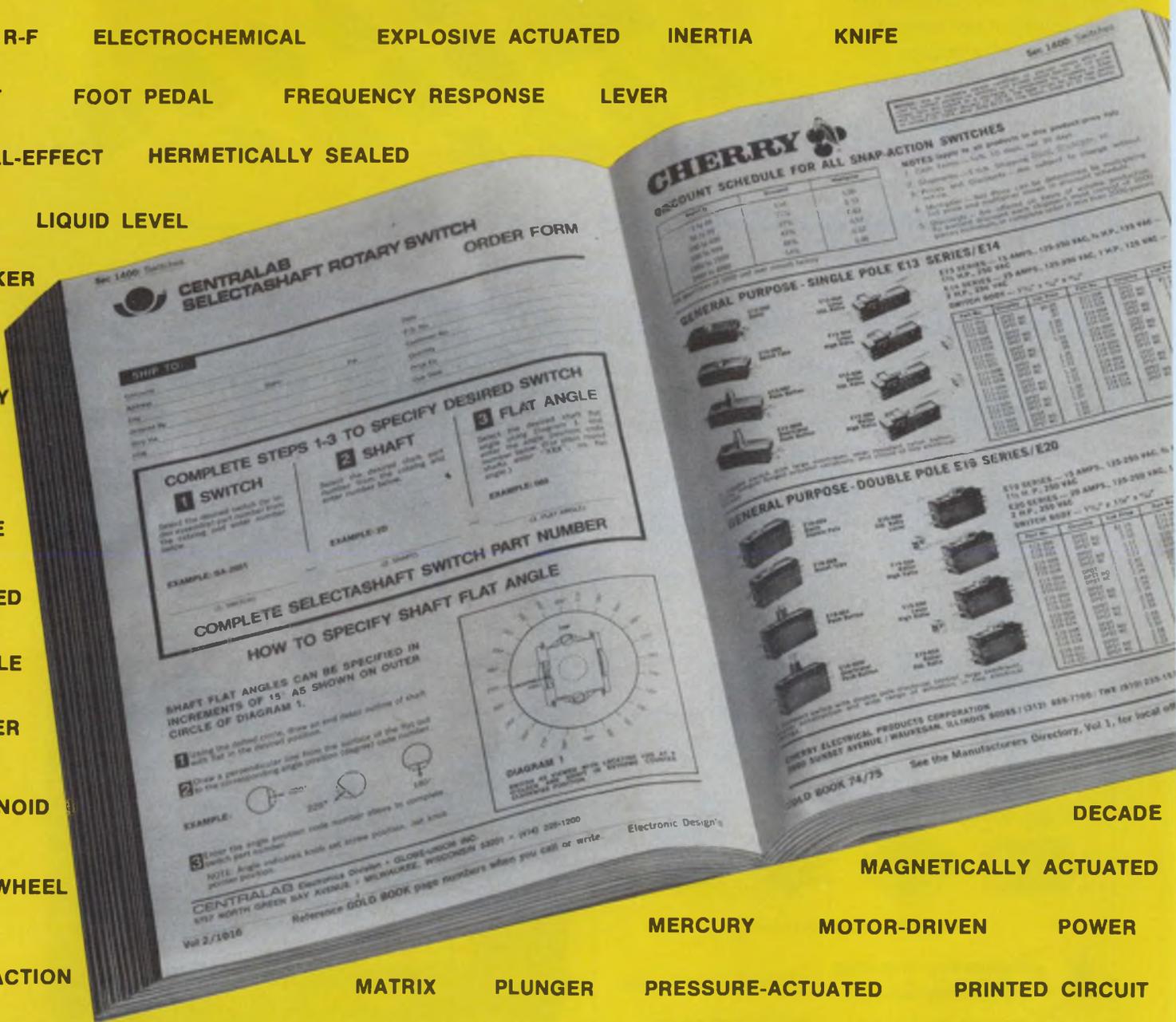
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INFORMATION RETRIEVAL NUMBER 130

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More than 52,500 of *Electronic Design's* qualified subscribers recommend or specify switches & relays for purchase. To aid in product search and selection, the PRODUCT DIRECTORY of *Electronic Design's* GOLD BOOK lists manufacturers of switches & relays in 95 sub-categories. For example there are 53 classifications of switches ranging from *Acceleration* through *Waveguide*; 42 classifications of relays from *Capacitance* through *Voltage Band*. To make it easier for you or your secretary to contact many manufacturers about a given item, each supplier's name is followed by *complete* street address, city, state, zip and phone *every time it appears*—no need to refer to other directories for missing portions of address.

CATALOG PAGES Volume 2 of *Electronic Design's* GOLD BOOK contains 235 catalog pages on *Switches; Relays & Solenoids*. You'll find most major manufacturers represented; many units contain detailed specs and ordering information, configurations, etc. Check these pages carefully before you specify . . . you may find the exact product you need.



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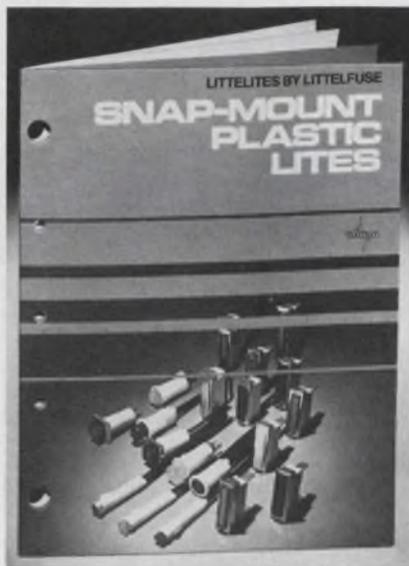
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Snap-mount lights

Three series of snap-mount plastic lights are described in an eight-page catalog. The catalog includes dimensional drawings and illustrations of each of the series. An easy-to-use ordering table facilitates correct designation of features, including termination, lamp, resistor, housing color and lens style and color. Littelfuse, Des Plaines, Ill.

CIRCLE NO. 369

Discrete semis and ICs

"Semiconductor Survey 1974," a 37-page short-form catalog, describes discrete semiconductors and ICs. The products are listed in application groups and include transistors and diodes (silicon and germanium), optoelectronic devices (emitters and sensors), digital and analog ICs and multi-chip components. Specifications and dimensional drawings are included. AEG-Telefunken, Englewood Cliffs, N.J.

CIRCLE NO. 370

Micromanipulators

Features of precision micromanipulators including motor-operated fine adjustment motions with micrometer readout to 0.01 mm are given in a five-page brochure. Hacker Instruments, Fairfield, N.J.

CIRCLE NO. 371

Wire-wound resistors

A revised and updated precision and power wire-wound resistor handbook includes a military style ordering guide and cross-reference chart, information on established reliability wire-wound resistors, a technical guide on resistance temperature characteristics and data on 200 resistor types. RCL Electronics, Irvington, N.J.

CIRCLE NO. 372

Open frame power supplies

The Omni series of open frame modular power supplies are described in a bulletin. Included are electrical specifications, mechanical dimensions, photos and pricing for single and multiple output models. Deltron, North Wales, Pa.

CIRCLE NO. 373

Stock and custom tool kits

Tool kits designed for maintenance, assembly and repair in the electronic, electrical, vending, computer, relay, appliance and office machine industries are featured in a 12-page catalog. Electronic Tool, Mount Vernon, N.Y.

CIRCLE NO. 374

Batteries

Nickel-cadmium cells and batteries are described in two six-page brochures. Gulton Battery, Metuchen, N.J.

CIRCLE NO. 375

Calculator handbook

To extend the usefulness of the HP-35 pocket-sized scientific calculator, a \$10 handbook gives the most efficient keystroke sequences for solving commonly encountered mathematical problems. The 169-page book has seven major sections dealing in detail with calculator operations, number theory and algebra, geometry and trigonometry, statistics, numerical methods and finance. Hewlett-Packard, 1501 Page Mill Rd., Palo Alto, Calif. 94304.

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INFORMATION RETRIEVAL NUMBER 132

This new 10-watt amplifier is tough -- even in a mismatch!

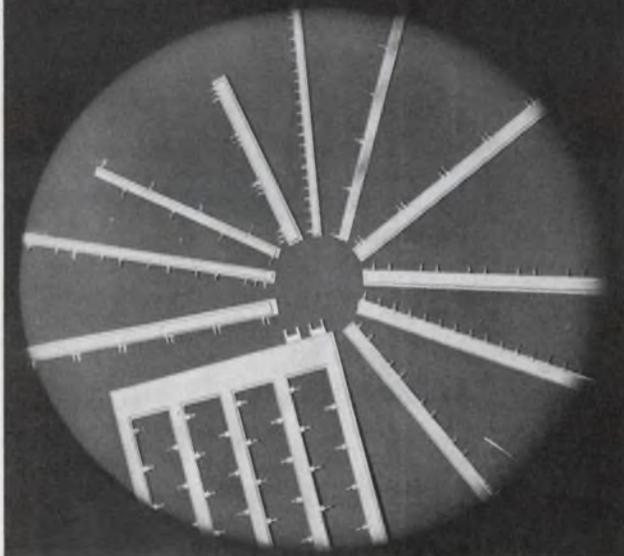


The Model 10LA is one broadband amplifier you don't have to treat with kid gloves. It will stand up under any mismatched load and provide you with 10 watts of swept power from 1-110 MHz. A directional wattmeter enables you to determine actual power delivered to the load. You can perform antenna and component testing, equipment calibration, NMR research -- any number of tests -- with complete confidence. It takes more than a mismatch to knock out our 10LA. Find out for yourself, write: Amplifier Research, 160 School House Road, Souderton, Pa. 18964. Phone: 215-723-8181



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INFORMATION RETRIEVAL NUMBER 133

ELECTRONIC DESIGN 20, September 27, 1974

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SMC SERIES

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INFORMATION RETRIEVAL NUMBER 210

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Our MINI-SLIDE switches spell quality and cost you less . . . Simply because we make 'em by the millions. Improved actuator coiled spring provides a smooth tactile operation and a contact wiping action. Available either for PC or hand wiring with or without mounting tabs.



SLS-220
SLS-120

ALCOSWITCH®

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INFORMATION RETRIEVAL NUMBER 211

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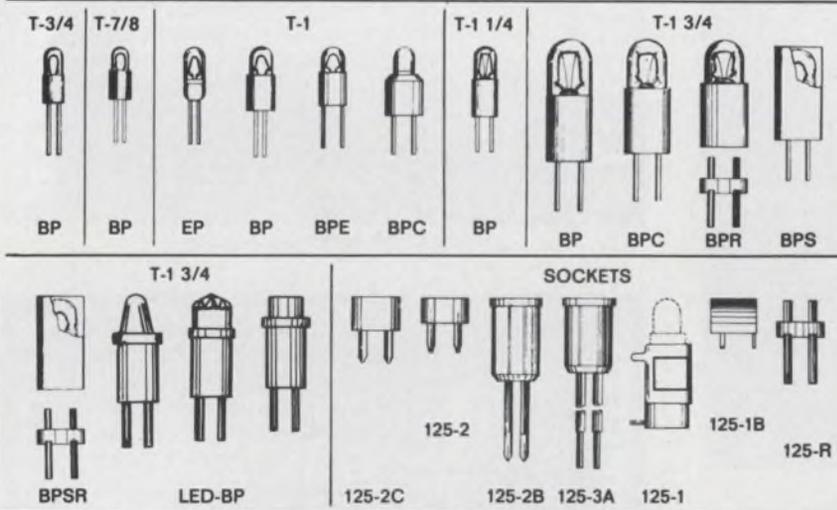
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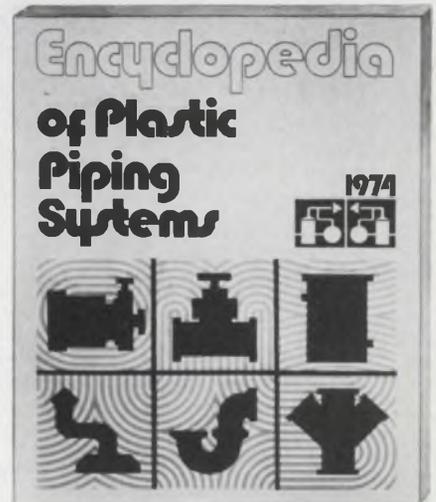
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OSHINO ELECTRIC LAMP WORKS, LTD. TOKYO, JAPAN

INFORMATION RETRIEVAL NUMBER 137

NEW LITERATURE



Plastic piping systems

A 240-page reference book for plastic piping systems contains data to enable users to design, estimate and purchase either partial or complete plastic air or fluid handling systems. Plastic Piping Systems, South Plainfield, N.J.

CIRCLE NO. 376

MOS LSI circuits

RAMs, ROMs, keyboard encoders, shift registers, standard patterns and calculator circuits are described and illustrated in an 18-page catalog. Electronic Arrays, Mountain View, Calif.

CIRCLE NO. 377

Optoelectronic components

LED displays, discrete indicators, optoisolators and solid-state relays are itemized in a short-form catalog. Electrical, optical and mechanical data are included. Monsanto, Palo Alto, Calif.

CIRCLE NO. 378

Cassette recorders

"Handbook on Incremental Digital Tape Cassette Recorders" is aimed at acquainting potential users with the advantages of incremental recording using magnetic tape cassettes as the storage medium. Low power, high-density data logging recorders are described as well as serial, parallel (TTL and CMOS levels) and RS 232C/TTY interfaces in OEM and stand-alone instrument configurations. Memorydyne, Newton Upper Falls, Mass.

CIRCLE NO. 379

Microwave ovens

A guide to microwave ovens, with tube complement, lists brand name, manufacturer and model number of all of the microwave ovens sold in the U.S. within the last 10 years. With each model number is the tube type number installed by the manufacturer, together with the type of the company's magnetron which is a direct, interchangeable equivalent. Amperex Electronic, Hicksville, N.Y.

CIRCLE NO. 380

Resistor networks

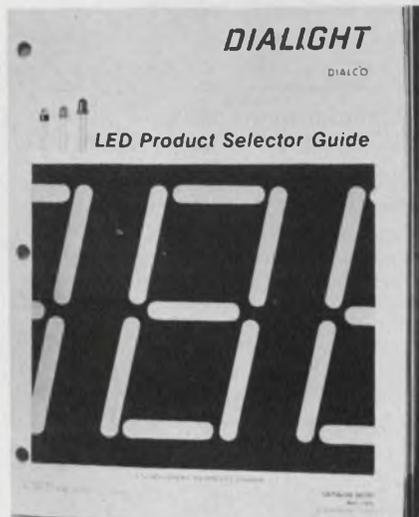
A 12-page resistor network bulletin covers 16 DIP networks (dual inline packaged) and two flatpack networks that solve resistor requirements for pull-up, pull-down, line termination, LED current limiting, pulse squaring networks and sense amp termination. Over 31 schematics are included. Beckman Instruments, Helipot Div., Fullerton, Calif.

CIRCLE NO. 381

Power supplies

Power supplies for IC logic and op amps are described in a four-page brochure. Acopian, Easton, Pa.

CIRCLE NO. 382



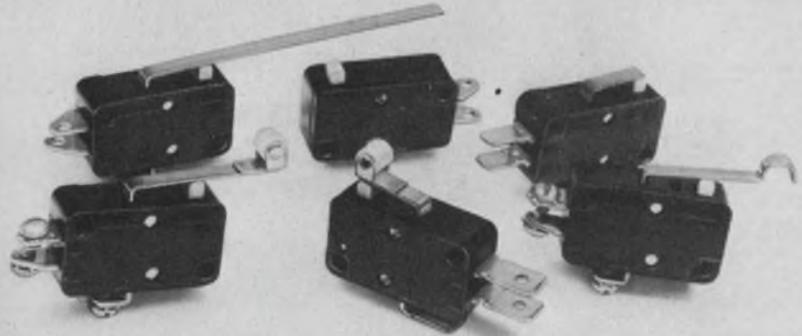
LEDs

A 72-page "LED Product Selector Guide" details the company's light-emitting diodes, indicators and related readout displays including red, green and yellow LED products. Dialight, Brooklyn, N.Y.

CIRCLE NO. 383

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INFORMATION RETRIEVAL NUMBER 138

NOBODY WANTS A CLUTTERBUG

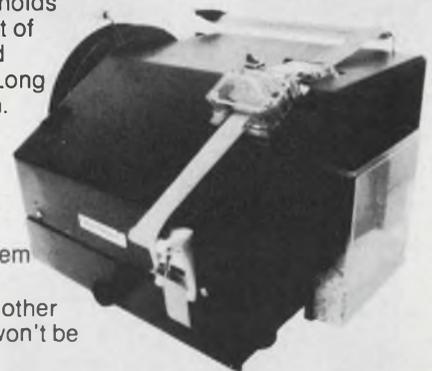
We had your end-user in mind when we engineered our compact, self-contained, no fuss, no muss, 50/60 cps, electro-mechanical punches.

With conveniences that show at the installation. Like bi-directional feed. An over-sized chad box that can contain chad from up to 1000 feet of punched tape without emptying. A drawer which holds a standard 8½" tape supply reel. An out of tape/tape tension switch. Side mounted take-up reel Operation "cool" at 100°F. Long life. Extremely low operator intervention.

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Model 518
Paper Tape Punch Station



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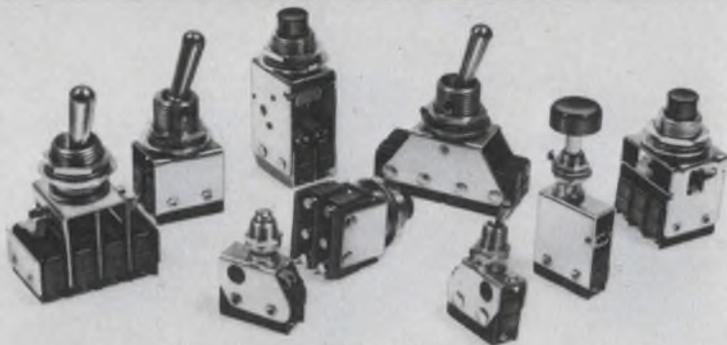
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Model FRK-L, Model FRK-H, Model FRT

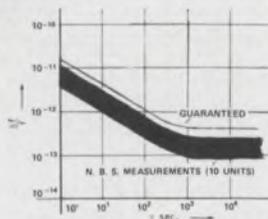
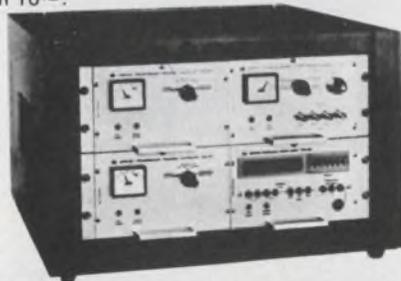
Rb-Atomic Resonance
Controlled Oscillator
Excellent Stability
10 min. Warm up
5 Year Lamp/Cell Warrantee



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Replaces PRIMARY FREQUENCY STANDARD, when coupled to Model EFR and synchronized to National Standard Frequency. Absolute Frequency is few parts in 10^{12} .



Model EFR Receiver-Controller
VLF/LF Frequency Coverage
Controls Frequency Standards
Time Synchronization Output for Clocks
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Model EDU Digital Clock
BCD - Outputs - Time
Leap Second Insertion
Price \$1,790.00

FOR EUROPE: Efratom Elektronik GmbH, 8000 Muenchen 90, Langobardenstr. 7 West Germany. Ph. (089) 647138

INFORMATION RETRIEVAL NUMBER 141

bulletin board

The Digital Products Div. of Fairchild has introduced a proprietary line of low-power Schottky TTL circuits (series 9LS), which makes it possible to produce 2-mW-per-gate digital systems that operate at twice the speed attained by standard TTL logic.

CIRCLE NO. 384

A family of TTL/MSI function devices built using advanced Schottky technology is now available in production quantities from Advanced Micro Devices. These devices, which include registers, decoders, multiplexers, counters and operators, function at approximately twice the speed of their standard TTL/MSI counterparts. All units that carry the 54S/74S series numbers are pin-compatible with like devices from Texas Instruments.

CIRCLE NO. 385

General Automation has announced a new combination of computer hardware and software, which has been tailored and priced to meet the needs of those users requiring remote job entry to an IBM 360/370 HASP system.

CIRCLE NO. 386

Tektronix' new software package, the PLOT-10/APL GRAPH-II, expands the potential of interactive graphics and APL with the Tektronix 4013 and 4015 computer display terminals.

CIRCLE NO. 387

Siltek International has added 20 new device types to its family of CMOS 4000 Series integrated circuits.

CIRCLE NO. 388

Signetics, 811 E. Arques Ave., Sunnyvale, Calif. 94086 has announced that two high-speed RAMs, formerly available in silicone packages, are now being manufactured in epoxy packages at \$14 (100). The price of the RAMs in silicone was \$28. The silicone RAMs have been superseded by the epoxy versions.

INQUIRE DIRECT

Stewart-Warner Microcircuits has introduced a line of **CMOS integrated circuits** in the standard 4000 series.

CIRCLE NO. 389

SE Labs (EMI) Ltd.'s low-cost **SM272 FRA transfer function analyzer/frequency response analyzer** is now offered with a digital readout as an option to standard analog meters.

CIRCLE NO. 390

Price reductions

A 33% price reduction of **semiconductor memories** and a 30% reduction of the **V-70 series memory management system** (memory map) has been announced by **Varian Data Machines**.

CIRCLE NO. 391

Litronix has announced a 14 to 22% price reduction on 13 **optoisolators** that use **LED emitters** and **npn phototransistors** or **photo-Darlington detectors**.

CIRCLE NO. 392

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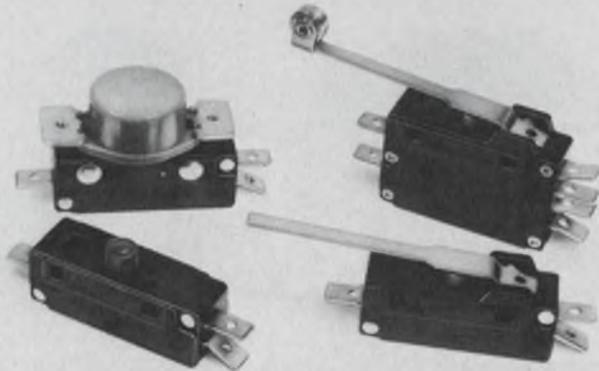
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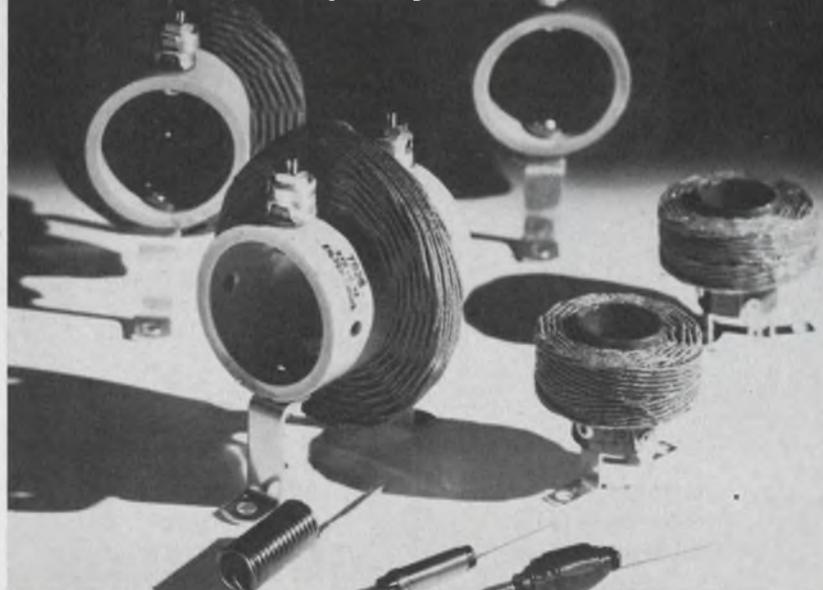
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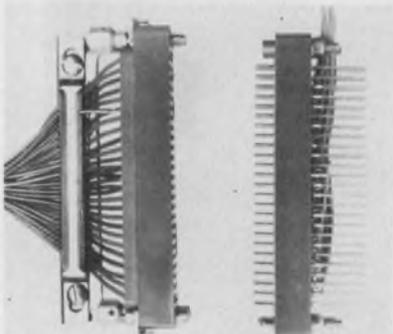
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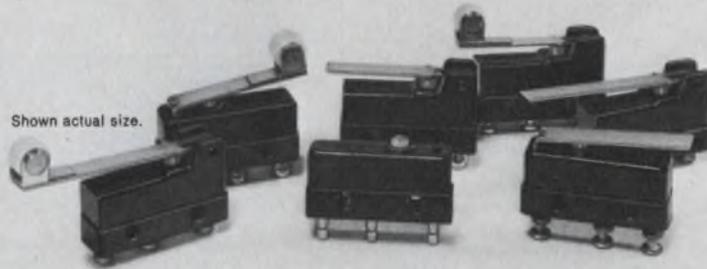
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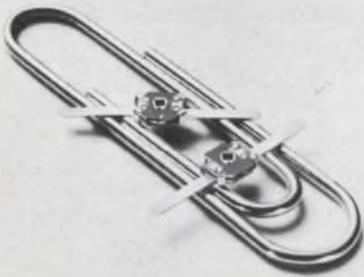
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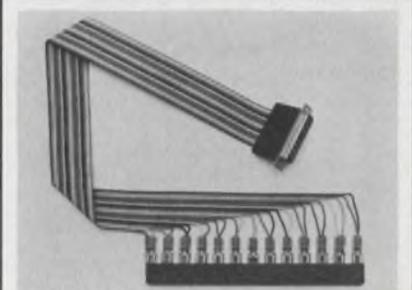
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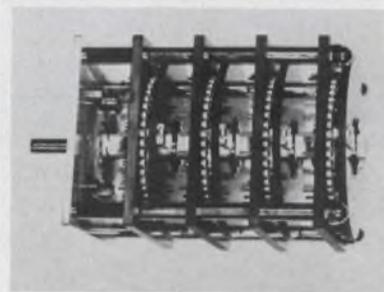
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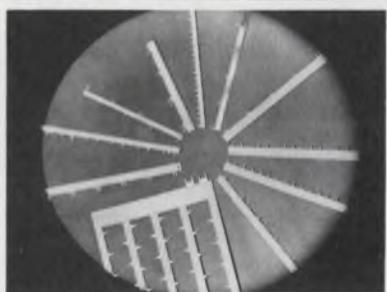
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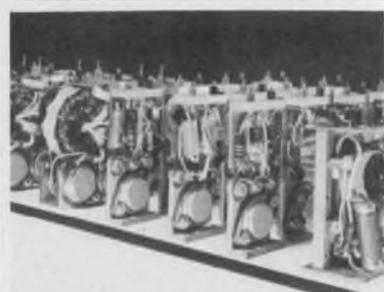
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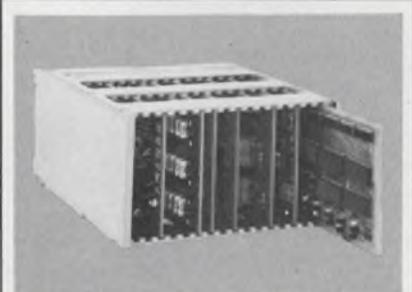
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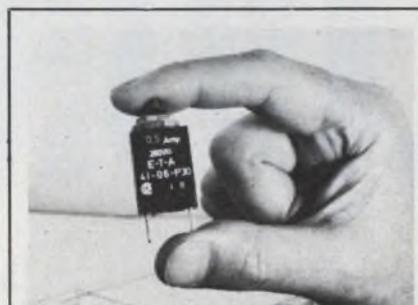
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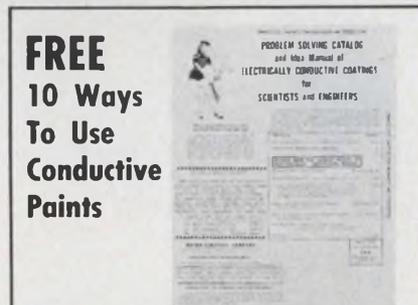
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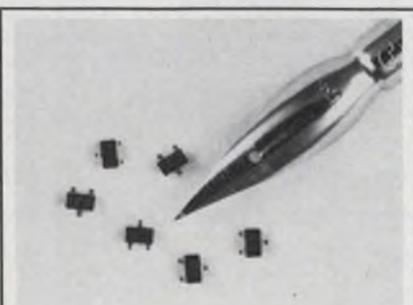
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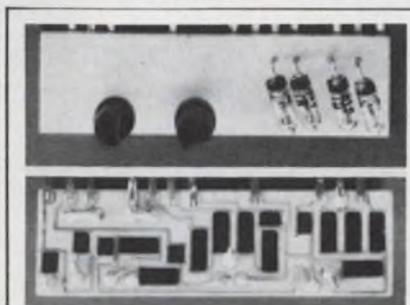
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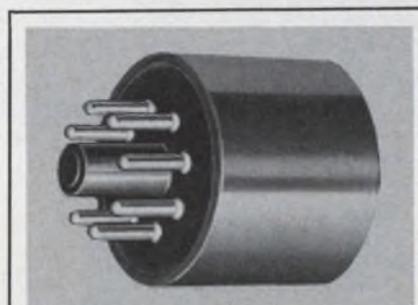
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■ WHO MAKES THEM

You'll find 97 manufacturers of power semiconductors beginning on pages 353, 428 and 457 of *Electronic Design's* GOLD BOOK (Volume 1—Product Directory). For your convenience, each manufacturer is listed with complete street address, city, state, zip and phone number.

■ SALES OFFICES—REPS DISTRIBUTORS

To find information about each power semiconductor manufacturer, turn to the *Manufacturers Directory*. Whenever possible it lists names of key officials, sales offices, export offices, foreign offices, U.S. and foreign reps followed by a list of U.S. distributors. In many cases there is additional data, when provided by the company: TWX, TELEX, cable address, facsimile equipment (make and call number), 800 (toll-free numbers) as well as number of engineers, number of employees and financial data.

■ CATALOG PAGES

Here's a rundown of the catalog pages you'll find on power transistors in *Electronic Design's* GOLD BOOK.

| Manufacturer | Starts in Vol. 2, Page: | Manufacturer | Starts in Vol. 2, Page: |
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FOR COOLING EQUIPMENT

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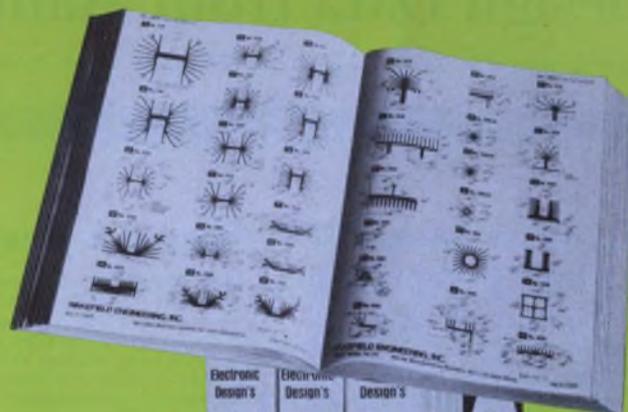
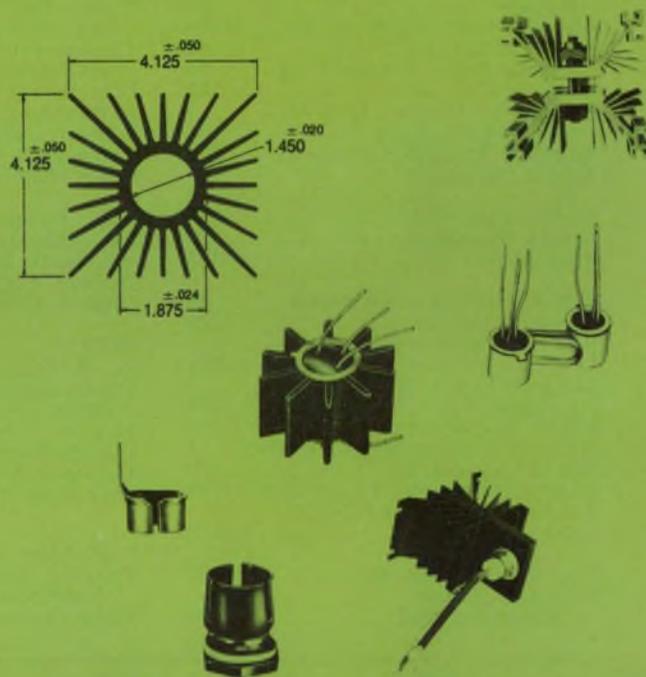
You'll find eleven categories of cooling equipment and related products listed in the *Product Directory of Electronic Design's GOLD BOOK*. For blowers and fans 38 manufacturers are listed. For thermoelectric cooling/heating modules, 17; circulating liquid cooling units, 21; heat sinks and dissipators, 50; thermal conductive coatings, 23; insulators and insulating hardware, 52; transistor mounting pads, 23; epoxy potting compounds, 41; silicone greases, 17; and washers, 18. As with power semiconductors, data about each manufacturer, his reps and distributors can be found by referring to the *Manufacturers Directory*.

■ HERE'S A RUNDOWN OF THE CATALOG PAGES YOU'LL FIND ON HEAT SINKS AND DISSIPATORS ALONE

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| Intl. Elec. Research | 4 | Vol. 2, 1291 |
| Intl. Rectifier | 2 | Vol. 2, 458 |
| Jermyn | 1 | Vol. 2, 1295 |
| Thermalloy Inc. | 1 | Vol. 2, 1304 |
| Unitrack | 1 | Vol. 3, 271 |
| Wakefield Engr. Inc. | 42 | Vol. 2, 1307 |

■ TAKE WAKEFIELD ENGINEERING, FOR EXAMPLE

Wakefield Engineering's 42 page catalog of semiconductor cooling products begins on page 1307 of Vol. 2. In addition to detailed specs and information about its heat sinks and thermal cooling products (most show curves of natural and forced convection characteristics) the pages include diagrams of 99 heat sink extrusion shapes with dimensions, surface area, and thermal characteristics. Wakefield's unit also provides guides on HOW TO SELECT HEAT SINKS, ENGINEERING DATA, HEAT SINK MOUNTING SPECIFICATIONS, HOLE PATTERNS, U.S. DISTRIBUTORS, EUROPEAN SALES AGENTS, U.S. FIELD SALES ENGINEERS, and lists other available Wakefield catalogs.



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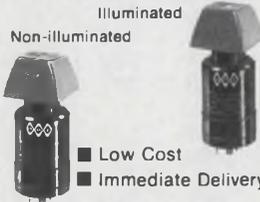
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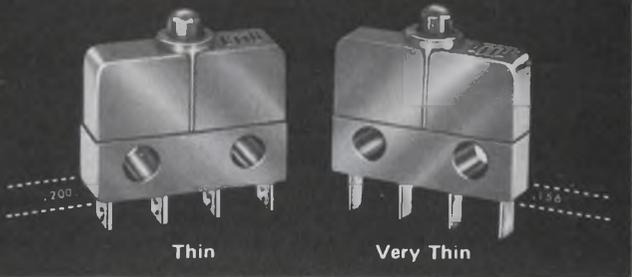
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