JULY - 1951 PRICE 75 CENTS

M C G R A W - H I L L P U B L I C A T I Q N

TESTING AIRCRAFT FUEL GAGES



Linear Standard Units...

THE ULTIMATE IN QUALITY. . .

UTC Linear Standard Audio Transformers represent the closest approach to the ideal component from the standpoint of uniform frequency response, low wave form distortion, high efficiency, thorough shielding and utmost dependability.

UTC Linear Standard Transformers feature . . .

- True Hum Balancing Coil Structure . . . maximum neutralization of stray fields.
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- Reversible Mounting . . . permits above chassis or sub-chassis wiring.
- Alloy Shields..., maximum shielding from inductive pickup.
- Hiperm-Alloy . . . a stable, high permeability nicket-iron core material.
- Semi-Toroidal Multiple Coil Structure... minimum distributed capacity and leakage reactance.
- Precision Winding... accuracy of winding .1%, perfect balance of inductance and capacity; exact impedance reflection.
- High Fidelity . . . UTC Linear Standard Transformers are the anly audio units with a guaranteed uniform response of ± 1 DB from 20:20,000 cycles.

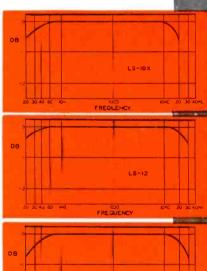
TYPICAL LS LOW LEVEL TRANSFORMERS

Type Nq.	Application	Primary Impedance	Secondary Impedance	±1 db from	Max. Level	Relative hum- pickup reduction	Max. Unbal- anced DC in prim'y	List Price
LS-10	Low impedance mike, pickup, or multiple line to grid	50, 125, 200, 250, 333, 500/ 600 ohms	60,000 ohms in two sections	20-20,000	+15 DB	—74 DB	5 MA	\$25.00
LS-I0X	As Above	As above	50,000 ohms	20-20,000	+14 DB	92 DB	5 MA	32.00
LS-12	Low impedance mike, pickup, or multiple line to push pull grids	50, 125, 200, 250, 333, 500/ 600 ohms	120,000 ohms overall, in two sections	120,000 ohms 20-20,000 overall, in two		74 DB	5 MA	28.00
LS-12X	As above	As above	80,000 ohins overall, in two sections	20-20,000	+14 DB	92 DB	5 MA	35.00
LS-26	Bridging line to single or push pull grids	5,000 ohms	60,000 ohms in two sections			—74 DB	0 MA	25.00
LS-19	Single plate to push pull grids like 2A3, 6L6, 300A. Split secondary	15,000 ohms	95,000 ohms; 1.25:1 each side	20-20,000	+17 DB	50 DB	0 MA	24.00
LS-21	Single plate to push pull grids. Split primary and secondary	15,000 ohms	135,000 ohms; turn ratio 3:1 overall	20-20,000	+14 DB	-74 DB	0 MA	24.00
LS-22	Push pull plates to push pull grids. Split primary and secondary	30,000 ohms plate to plate	80,000 ohms; turn ratio 1.6:1 overall	20-20,000	+26 DB	—50 DB	.25 MA	31.00
LS-30	Mixing, low impedance mike, pickup, or multi- ple line to multiple line	50, 125, 200, 250, 333, 500/ 600 ohms	50, 125, 200, 250, 333, 500/600 ohms	20-20,000	+17 DB	—74 DB	5 MA	25.00
LS-30X	As above	As above	As above	20-20,000	+15 DB	—92 DB	3 MA	32.00
LS-27	Single plate to multiple line	15,000 ohms	50, 125, 200, 250, 333, 500/600 ohms	30-12,000 cycles	+20 DB	—74 DB	8 MA	24.00
LS-50	Single plate to multiple line	15,000 ohms	50, 125, 200, 250, 333, 500/600 ohms	20-20,000	+17 DB	—74 DB	0 MA	24.00
LS-51	Push pull low level plates to multiple line	30,000 ohms plate to plate	50, 125, 200, 250, 333, 500/600 ohms	20-20,000	+20 DB	—74 DB	1 MA	24.00
LS-141	Three sets of balanced windings for hybrid ser- vice, centertapped	500/600 ohms	500/600 ohms	30-12,000	+10 DB	—74 DB	0 MA	28.00

TYPICAL LS OUTPUT TRANSFORMERS

Type No.	Primary will match following typical tubes	Primary Impedance	Secondary Impedance	±1 db from	Max. Level	List Price
LS-52	Push pull 245, 250, 6V6, 42 or 2A5 A prime	8,000 ohms	500, 333, 250, 200, 125, 50, 30, 20, 15, 10, 7.5, 5, 2.5, 1.2	25-20,000	15 watts	\$28.00
LS-55	Pušh pull 2A3's, 6A5G's, 300A's, 275A's, 6A3's, 6L6's	5,000 ohms plate to plate and 3,000 ohms plate to plate	500, 333, 250, 200, 125, 50, 30, 20, 15, 10, 7.5, 5, 2.5, 1.2	25,20,000	20 watts	28.00
LS-57	Same as above	5,000 ohms plate to plate and 3,000 ohms plate to plate	30, 20, 15, 10, 7.5, 5, 2.5, 1.2	25-20,000	20 watts	20.00
LS-58	Pust, pull parallel 2A3's, 6A5G's, 300A's, 6A3's	2,500 ohms plate to plate and 1,500 ohms plate to plate	500, 333, 250, 200, 125, 50, 30, 20, 15, 10, 7.5, 5, 2,5, 1.2	25-20,000	40 watts	50.00
LS-6LI	Push pull 6L6's self bias	9,000 ohms plate to plate	500, 333, 250, 200, 125, 50, 30, 20, 15, 10, 7.5, 5, 2.5, 1.2	25.20,000	30 watts	42.00







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electronics



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July • 1951

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July, 1951

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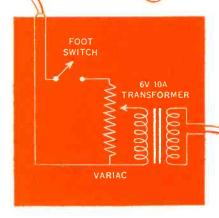
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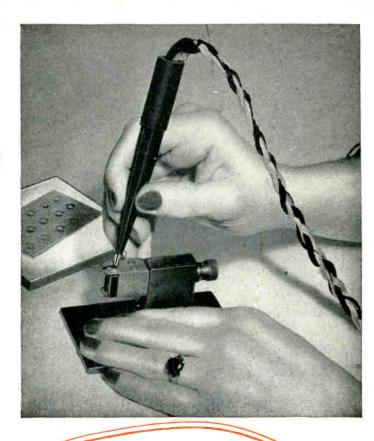
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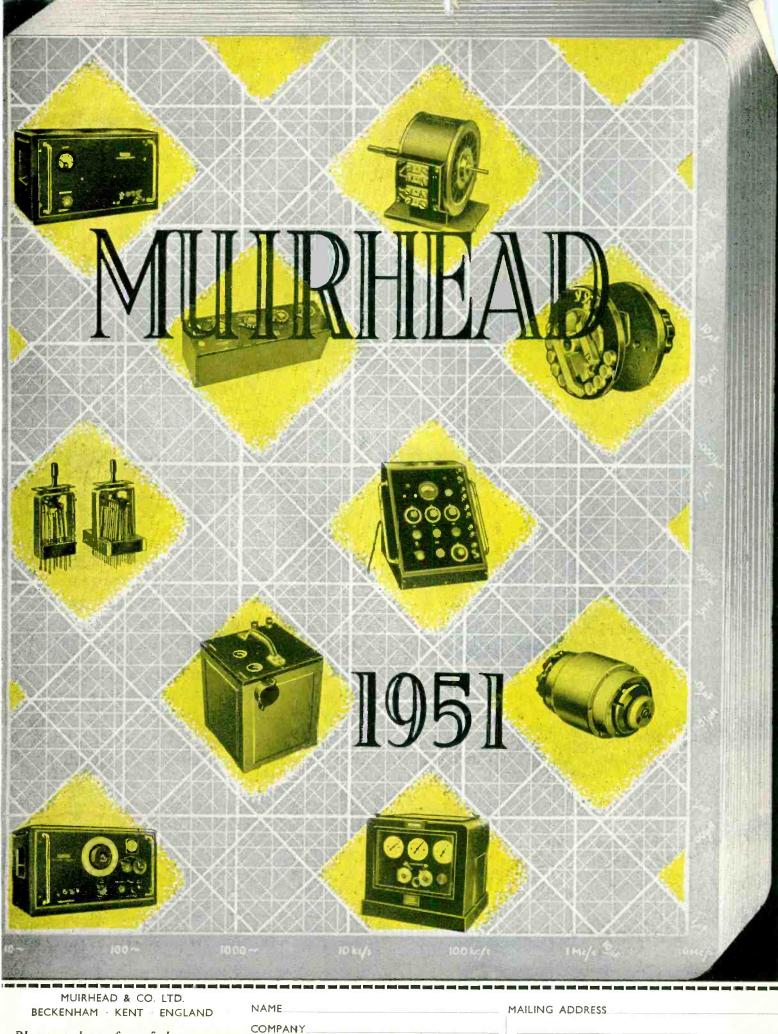
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, separated and insulated by

NOT FOR SALE—You can make electric soldering tweezers of your own. Those shown here have standard "MM" blades, separated and insulated by bakelite. They are assembled and wired as illustrated. By varying the voltage, the tip cross-section, and the contact area, you can adapt this tool to your specific application.

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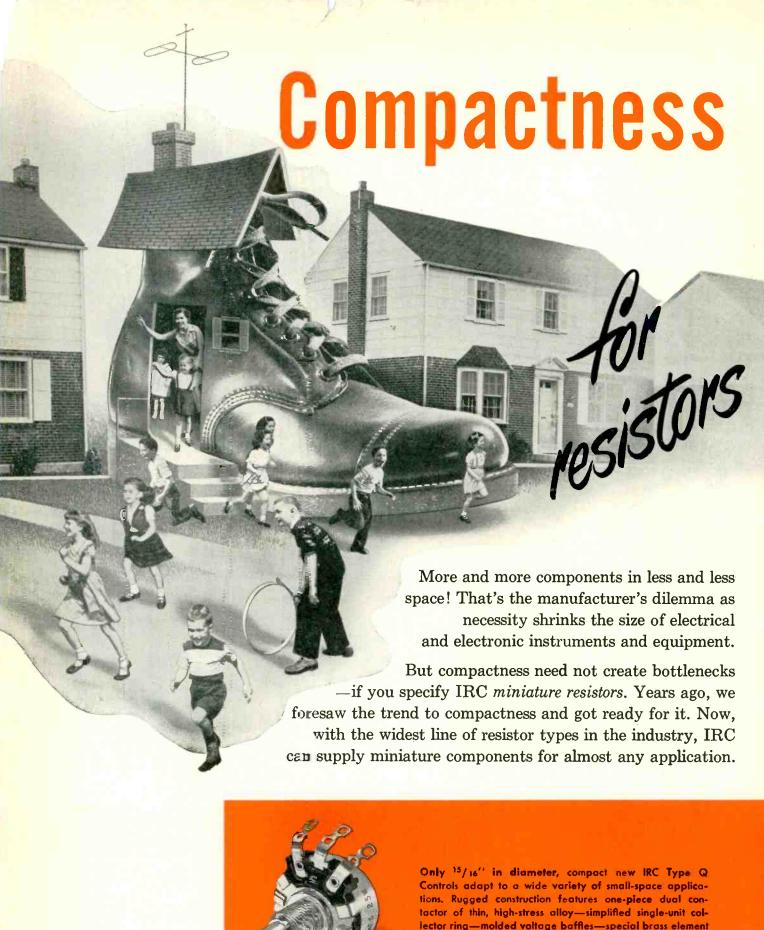
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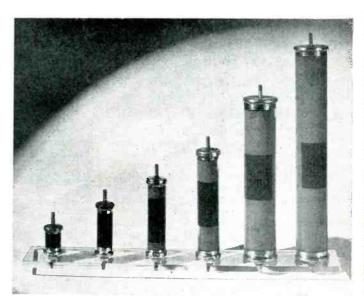
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MM	inforced phenolic laminates and similar materials, yet competitive in price.	Rigidity (Young's Mod. x 10")	1.87	2.84	3.38
		Impact Strength (Izod)	13.5	19.2	25.0
GLASTIC	Similar to Glastic MM but made with a resin of greater strength and flame	Tensile Strength (psi)	11,100	17,300	44,500
GF	resistance; well suited to resist pound- ing, required in driving slot wedges.	Compressive Strength (psi)	33,500	53,100	57,500
		Hardness (Rockwell M)	90	90	90
GLASTIC GW	Premium grade, much greater strength than either MM or GF; 45,000 psi at 150°C after 200 hours; requires sharp carbide tools in fabricating.	Specific Gravity	1.59	1.51	1.75
		Water Absorption (% ASTM)	.54	.36	.61
		Arc Resistance (ASTM)	150	60	120
GLASTIC	Similar to GW but uses unidirectional glass cloth for maximum tensile	Dielectric Strength (V/M)	300	280	210
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GLASTIC A	Tough and thin; will crease without serious loss in dielectric (500-700	Flexural Strength (psi)	25,200	32,200	45,000
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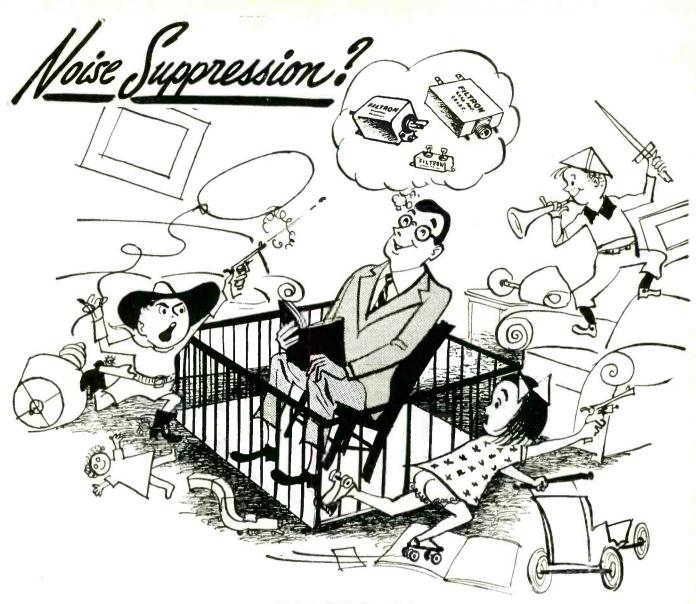
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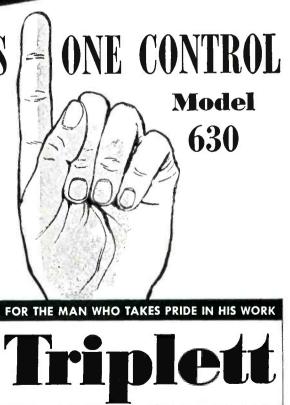
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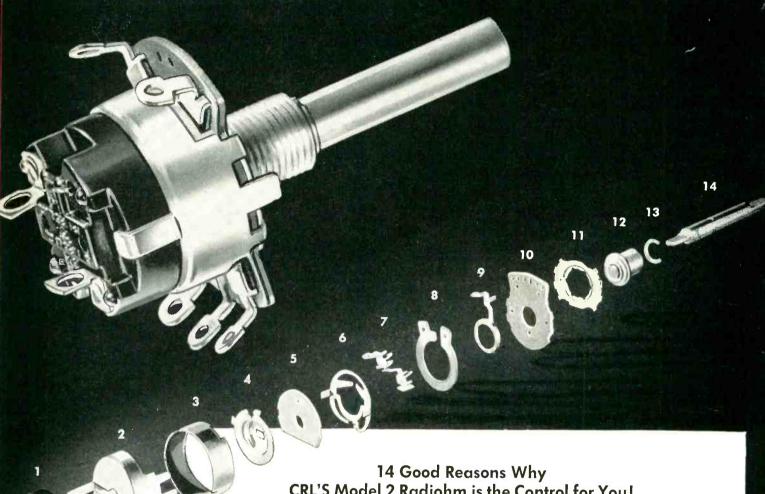
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- 3. Contact spring: Phosphor-bronze. Double wiping. Uniform mechanical pressure and positive electrical contact on resistor element and contact ring assures noiseless action.
- 4. Shoe: High grade, laminated phenolic. High insulation resistance in humidity.
- 5. Center terminal: Cadmium tipped, non-oxidizing. Easily soldered.
- 6. Terminals: Clinch type, tightly crimped to resistor element for mechanical rigidity and electrical continuity when soldered or in
- 7. Ground plate: Cadmium plated steel
- 8. Bushing: Cadmium plated steel, Accurately machined for minimum shaft-bushing clearance, smooth rotation.

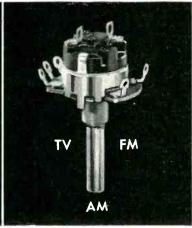
- Shoft: Many variations meet diversified application. Round, Flatted, Slotted, Split knurl, Finger tip knurl. Also nylon for high voltage application.
- 10. Cover: Cadmium plated steel, completely shields resistor element.
- 11. Switch: Directly operated from shaft. Positive on and off action. Standard, high and low current ratings in varied combinations.
- 12. Switch terminals: Ends shaped for wraparound leads. Hot tin-dipped for easy soldering. Mechanically anchored - will not loosen during soldering operations. Surfaces elevated; cannot short to cover legs.
- 13. Insulator: High dielectric strength permits breakdown rating of 1,000 volts, R.M.S.
- 14. Variety of unit types: Single or twin, concentric shaft twins, plain or switch type, any of these tapped or untapped. Simple design permits wide variation at minimum

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New Centralah Model 2 Radiohm Control — Left, single unit plain type, untapped; right, twin unit plain type, untapped. Both with single shafts.

New Centralah Model 2 Radiohm Control—control shown is a single unit switch type, tapped. Control has single shaft.

New Centralab Model 2 Radiohm Control — this control is a twin unit switch type, untapped. It has a single shaft.

New Centralab Model 2 Radiohm Control — Left, twin unit plain type, front section tapped; right, twin unit switch type, rear section tapped. Both units have concentric shafts.

OF SPECIAL INTEREST....

The World's Smallest Volume Control — Centralab's Model 1





Smaller than a dime! Less than ¼" thick.

Available in plain or switch type.

Ideal for miniature electronic gear.

FOR COMPLETE SPECIFICATIONS AND SAMPLES

If you want further information — taper curves, physical dimensions, engineering specifications on Centralab's Model 2 Radiohms or Model 1's, mail in coupon below. *Manufacturer's samples on request*.



Division of GLOBE-UNION INC. • Milwaukee

Centralab Div. Globe-Union Inc. 914 E. Keefe Ave., Milwaukee 1, Wis.

Please send me Control bulletin No. 42-85 on Centralab Model 2 Radiohms.

Also include Control bulletin No. 42-19 on Centralab's Model 1 miniature control.

Name	Title	

Company Address.

www.americanradiohistory.co

For Negative Resistance-Voltage Characteristics

GLOBAR

TYPE BNR RESISTORS

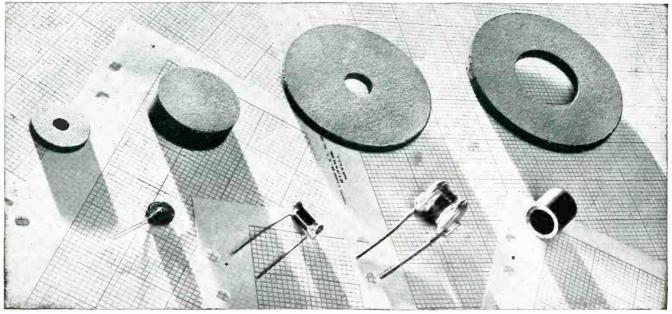
Typical applications where these resistors operate successfully include:

- Small motors to prevent arcing of governor contact points.
- 2 Stabilizing rectifier circuits by limiting peak voltages.
- Woltage control circuits in electronic devices.
- 4 Protection of solenoids in direct current circuits.

Responding instantly to voltage changes, GLOBAR type BNR Silicon Carbide Resistors provide increased resistance as a potential is decreased. Conversely, as a potential is applied, resistance decreases. These resistors are what is commonly referred to as voltage sensitive. They are used to dampen the effect of transient voltages and provide instant protection for electrical circuits.



Bulletin GR2 contains useful engineering data on GLOBAR BNR Ceramic Resistors. Copies will be supplied immediately upon request. Write Dept. V-71, The Carborundum Company, GLOBAR Division, Niagara Falls, N. Y.



Resistors of this type are readily made to meet exact specifications. Working samples are available when necessary. To be sure of receiving resistors made to correct specifications, the following information should be furnished:

- a. Type of apparatus in which resistors are to be used.
- b. Method of mounting and space limitations.
- c. Normal operating voltage and peak voltage if available.
- d. Resistance and inductance of the circuit if available.
- e. Ohmic resistance of the resistor and allowable plus or minus tolerance.
- f. Maximum voltage applied continuously or intermittently.
- g. Duration of load and elapse of time between applications.

 Furnishing these data will also avoid unnecessary delay and confusion

Furnishing these data will also avoid unnecessary delay and confusion.



GLOBAR Ceramic Resistors BY CARBORUNDUM

"Carborundum" and "Globar" are registered trademarks which indicate manufacture by The Carborundum Company, Niagara, Falls, N. Y

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The new Marka-Sweep Model Video extends to video circuit design and testing the speed, accuracy and convenience that has attended the use of sweep generators at higher frequencies. The oscilloscope pattern shown above is an exact reproduction of the detected output of the Model Video.

FEATURES:

Sweep Method: All Electronic

Sweep Ranges: 50 kc to 10 mc,

50 kc to 20 mcs.

Frequency marks of the pip type: Crystal positioned at 1, 2, 5, 10, 15, 20 mcs. Other frequencies to special order.

Variable frequency pip mark formed by injecting CW from auxiliary signal generator (Kay Accessory Oscillator now available.)

Separate attenuators on markers and video.

Price: \$495 f.o.b. Factory. 10% higher outside U.S.A. and Canada.

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Pine Brook, New Jersey



Dr. William Gilbert 1544-1603

This great English physicist and physician to Queen Elizabeth is universally recognized as the "father of electricity." He first advanced the hypothesis that the earth itself is a great magnet, and confirmed his ideas by various magnetic experiments. The gilbert, C.G.S. unit of magnetomotive force, is named for him.



From an Original Drawing made for OHMITE

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OHMITE

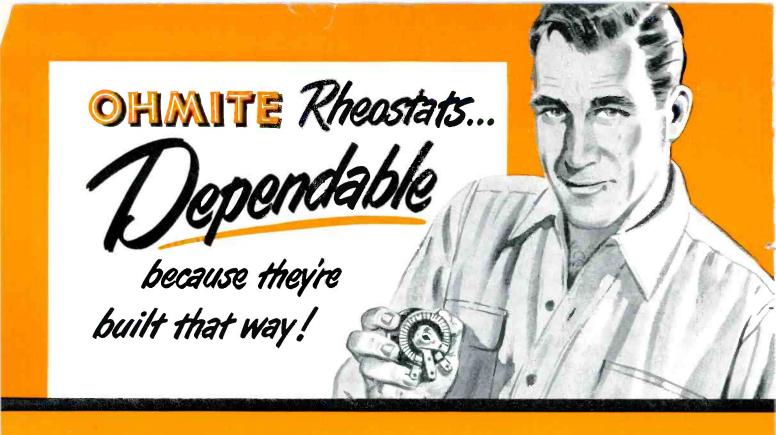
RHEOSTATS

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More manufacturers have mandardized on Ohmite rhecstats for their products... more companies are huming Ohmite rheostats for their own use... than any other make of rheostat on the market, today. The primary reason for this industry-wide preference for Ohmite rheostats is their proven dependability... their ability to give exira years of unfailing trouble free service.



METAL-GRAPHITE BRUSH

Perfect contact with negligible wear on the wire is insured by the metal-graphite contact brush (varied to fit the current and resistance) and the large, flat contact surface.

LARGE PIGTAIL SHUNT

Current is carried directly to the slip-ring by a pigtail shunt of ample size, assuring an uninterrupted connection at all times. Large slip-ring minimizes mechanical wear.

INSULATED SHAFT

High strength ceramic hub insulates the shaft and bushing from all live parts. Underwriters' Laboratories listed models are available.

VITREOUS ENAMEL BOND

Core and base are bonded together by vitreous enamel into one integral unit. Each turn of wire is also permanently locked in place by vitreous enamel

UNIFORM CONTACT PRESSURE

Tempered steel contact arm forms a long steel spring which assures uniform contact pressure. Universal joint action of the brush maintains "Aush-floating" contact.

ALL CERAMIC AND METAL CONSTRUCTION

Ohmite rheostats have a ceramic core, base, and driving hub. There are only ceramic and metal in their construction—nothing to char, burn, shrink, or deteriorate. They are designed for long, trouble-free life.

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WRITE on company letterhead for Catalog and Engineering Manual No. 40.

ZBe Right with...

OHMITE

RHEOSTATS · RESISTORS

TAP SWITCHES

Instrument NEWS



Spoilage Costs Cut 90% With G-E Roughness Scales

Air Maze Company, Cleveland, Ohio, manufacturers of filters for all purposes say that their G-E surface roughness scales and specimens cut spoilage costs 90% which otherwise would have increased manufacturing expenses. "They have provided us with real savings in both time and money," says H. W. Matlock, chief draftsman for Air Maze.

Daily use is made of the roughness specimens in preparing the specifications and blueprints for component filter parts. Then, with the handy lightweight scales, Air Maze makes certain that the finishes on parts received from their vendors are correct. When there is a question as to the suitability of a standard part, they end the argument.

G-E surface roughness scales and specimens are particularly adaptable when any parts, manufactured by several subcontractors or by different departments in one manufacturing plant, must have uniform surfaces roughness. The scales and specimens provide a common basis for the engineer, draftsman, machinist, and inspector in determining, supplying, and approving the correct surface finish.

Jet Engine Opposite-Polarity Ignition Coils are shown below being tested with a General Electric crest voltmeter. The crest voltmeter measures either positive or negative crest voltages. Portableit does not need an external power supply.



DINGS MAGNETIC SEPARATOR COMPANY RECOMMENDS USE OF G-E GAUSS METER



Dings Magnetic Separator Co., Milwaukee, Wis., manufacturer of magnetic separators, recommends the G-E gauss meter for determining the strength of the horseshoe magnets used in their separators. Reports Mr. Karl A. Blind, Chief Engineer at Dings, "We are recommending the G-E gauss meter to operating personnel in plants where magnetic separators are an essential part of production machinery. We believe operators should acquire and use the gauss meter for purposes of effectively and rapidly checking the full operating efficiency of the magnet."

Magnetic separators consist of either one or a series of permanent or electromagnets. Materials which contain magnetic foreign particles are passed in a steady flow near the magnets-which remove these foreign particles. This process sometimes requires much repetition, with the strength of the magnets varied each time. In order to do this, the exact strength of each magnet has to be controlled to within narrow limits. The General Electric gauss meter enables operators to maintain these limits.

Designed for unidirectional fluxes, the gauss meter can be used for checking flux densities and flux gradients in air gaps, and for measuring flux density in iron structures.



Ten Photoelectric Recorders Used at Battelle Institute



Battelle Institute, Columbus, Ohio, is now using ten General Electric photoelectric recorders in its industrialresearch laboratory.

Some of the recent applications for nal gh-2115

metallic and plastic substances. It is also used as a sensitive recording instrument for use in measurement and non-destructive testing where it responds to and signals from devices such as: ionization gages, photoelectric cells, thermocouples, transducers, and strain gages.

The Type CE photoelectric recorder measures d-c volts or amperes directly, and will record almost any quantity that can produce a d-c signal which varies in proportion to the quantity being measured. Chart speeds range from 1/2 inch per hour to 72 inches per minute; sensitivities are as high as 1.0 microampere full scale; response periods as fast as 1/5 second for full-scale deflection.

	which photoelectric recorders have be used at Battelle are: differential thern analysis, and recording short-time hig temperature creep qualities of various
GENERAL @	& ELECTRIC

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Plea	se send me the following bulletins:
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×	for planning an immediate project
	Gauss Meter (GEC-238A)
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	Type CE Photoelectric Recorder (GEC-254)
	Crest Voltmeter (GEC-380)
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	APANY
STR	ET
CITY	ZONE STATE



When the Weller Manufacturing Company, Easton, Pa., was completing the development of its new electric soldering gun, they were confronted with this problem: The 3/8" Revere Copper Rod used to replace the secondary coil in the transformer had to maintain its rigidity yet be sufficiently soft so that during the shearing, coining, and bending operations there would be no breaking or splitting of the rod.

Revere's Technical Advisory Service recommended a certain temper copper rod. It was discovered that Weller was getting a twist in the rod when it was installed in the assembled gun. Other tempers were tried

and tested. Then a copper rod of a slightly harder temper than the first was recommended. That was it! Proper temper was the key. Proper temper was also the key to the .291 dia. copper rod used for the Soldering tip itself. For this, too, had to retain its rigidity and yet remain soft enough to be coined, punched, and formed without fracture.

unit of increased capacity.

"In addition to being extremely helpful in arriving at the proper tempers, Revere also recommended that we specify our rod in multiple lengths, and thus save considerably on scrap. They were also helpful in solving the problem of attaching the brass sleeve to the secondary rod in

our Soldering Gun," the Weller Manufacturing Company tells us.

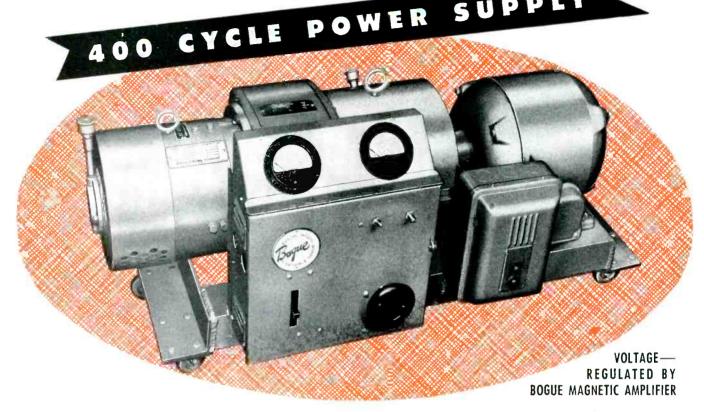
So you see, Revere's interest in your problem does not stop with the recommendation of its products. Perhaps Revere can help you. Why not take your current problem to the nearest Revere Sales Office and see?

REVERE

COPPER AND BRASS INCORPORATED

Founded by Paul Revere in 1801
230 Park Avenue, New York 17, New York

Mills: Baltimore, Md.; Chicago, Ill.; Detroit, Mich.; Los Angeles and Riverside, Calif.; New Bedford, Mass.; Rome, N. Y. Sales Offices in Principal Cities, Distributors Everywhere. The NEW Bogue Portable



This compact completely self-contained motor generator set operates from commercial 60 cycle power to furnish up to 5 KW of 400 cycle power for use in testing 400 cycle radio and aircraft equipment as well as operation of high speed machine tools, etc. The unit is complete with starting equipment and automatic voltage regulator to maintain the output voltage constant

over varying loads. It can be supplied with either single phase or three phase 60 cycle motor and either single or three phase 400 cycle generator. The output wave form is extremely pure, the total harmonic content of the three phase design being approximately 1%. In order to move the unit from one location to another casters are supplied as standard equipment.

HOW BOGUE SERVES INDUSTRY

Bogue Electric specializes in producing electrical equipment required for special applications such as pure 400 cycle power...less than 1% harmonics; low ripple dic generators for precision testing of electronic equipment; magnetic amplifiers for reliable, constant speed and voltage control. Bogue electronic engineers have solved hundreds of confidential problems. They are always available to talk over your specialized control or power problems, too!

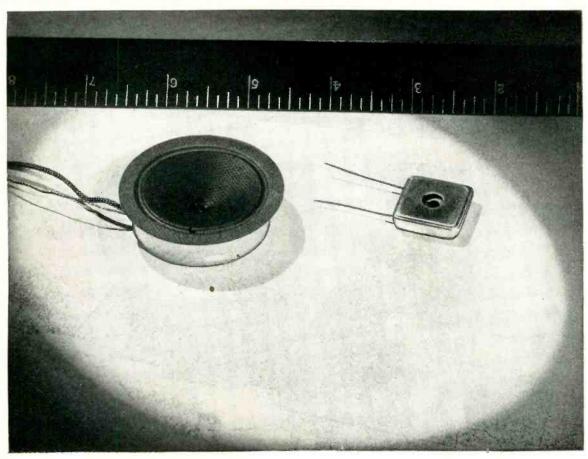
BOGUE ELECTRIC MANUFACTURING COMPANY

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MAGNETIC AMPLIFIERS . MAGNETIC CONTROLLERS . AC & DC MOTORS AC & DC GENERATORS . VOLTAGE & SPEED REGULATORS . SWITCHBOARDS





A Brush DEVELOPMENT

hearing aids have shrunk to half their size

Brush has been the leader in cutting the size of hearing-aid microphones. Now hearing aids are small, almost unobservable and, most important of all, do their job better. Deaf people no longer have to wear bulky, unsightly and hard-to-operate hearing aids.

The Brush Development Company for more than 21 years has studied the intricate, though powerful, piezoelectric effects of crystals. The principle of this smaller, more powerful microphone, as pioneered and developed by Brush, is that a tiny, jewel-like crystal converts sound waves into powerful electric signals. This is an example of continued engineering effort by Brush to make available to industry the many benefits and uses of piezoelectric materials, whether in hearing aids, measuring instruments or recording devices.

Manufacturers of

ACOUSTICAL EQUIPMENT

Microphones — General Purpose and Specialized Hearing Aid Microphones Earphones — Of Many Types

MAGNETIC RECORDING DEVICES

Tape
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Universal Strain Analyzers Surface Analyzers Multichannel Direct Writing Oscillographs AC, DC and Carrier-type Amplifiers Uniformity Analyzers Transient Recorders

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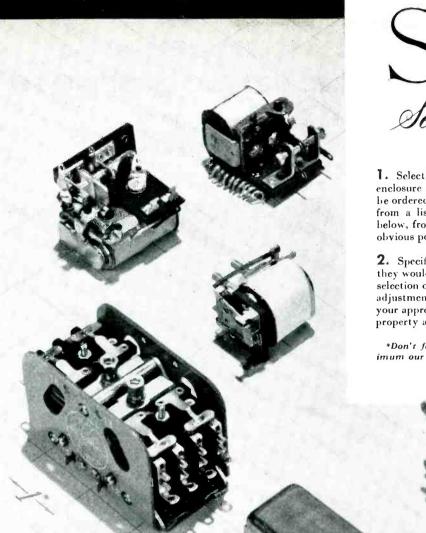
Generators and Analyzers Laboratory Equipment

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THE BRUSH DEVELOPMENT COMPANY

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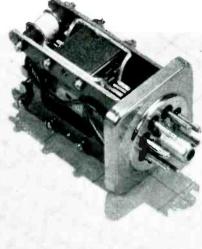


SIGMA Sensitive Relays

AVAILABLE SERVICE

- 1. Selection from six basic designs, each with many enclosure and mounting styles. Each such selection may be ordered with a coil resistance and adjustment chosen from a list of standard codes. The classification chart below, from the new catalog, calls attention to the most obvious points of comparison between design series.
- 2. Specification by our application engineers of the relay they would use if faced with your problem.* This includes selection of type and enclosure as well as design of special adjustment and electrical characteristics, all subject to your approval and test. The resulting specification is your property and will be disclosed only at your request.

*Don't fail to describe all of it, so as to reduce to a minimum our guesswork!



RELAY	SWITCH TYPE (MAX.	SWITCH		COIL	MAXIMUM COIL SPEED	MOUNTINGS, ENCLOSURES, AND DESIGNATIONS			PARTICULAR		
SERIES	CONTACT ARRANGE- MENT)	RATING	Maximum (Standard)	"Aircraft"	RESISTANCE OR IMPEDANCE	PATION (40°C rise)5	RANGE**	Open Frame	Dust Cover	Hermetic Seal	FEATURES
4	SPDT (1-C)	2 amp.	.02 watt DC	.04 watt DC	12,000 ohms	1.5 watts 85°C	002D15	4F — 🗱	4A - 4R -	4RJ —	Medium price, sensitivity end precision.
5	SPDT	2 amp.	,001 watt DC	.010 watt DC	16,000	2.0 watts	.004020 sec. Slow- release, 5F—	5F — 🔊		SRJ-	High sansitivity, shock re- sistance, stability, and eibre-
J	(1.C)	A	.007 watt AC, internal rectifier	.015 watt AC, internal rectifier	ohms	100°C	with slug, evailable.			510 -	tion resistance.
41	SPDT	2 amp.	.04 watt	Not generally recommended	12,000 ohms	1.5 watts 100 C	.002015 sec.	41F—	41R —	41RJ -	High AC sansitivity, form mechanical life; fast—good for pulsing; high load rat
41	(1-C)	styles 5 amp	5 amp U.I VA No	Not generally recommended	\$90,000 ohms	9.5 VA 100°C	Operates in 1st half cycle.	(other open mountings available)	41RO	4110	for pulting; high load rat ingo possible; fow price.
6	4PDT (4C)	2 amp. 5 amp. with SPDT DPDT	.012 watt -	- 4C contacts — 1C contacts DLARIZED	25,000 ohms (12000/12000 in dual coils)	3.0 watts 85-C	.00705 sec.	6FX 6FY 6FZ		6HX 6HY 6HZ	Polerizad: blesed, letchin, or 3-position (null-teaking) high gain (load power/operating power ratio); multicontect.
7	SPDT (1-C)	2 amp.	.0001 watt, I	switching, gen'l keying setting DLARIZED	12,000 ohms (3200/3200 in dual coils)	1.5 watts _85 °C	.001007 sec.		7AOX 7AOY 1AOZ	7JOX 7JOY 7JOZ	Polarized: biased, letchin or 3-position [null-seating For teletype and taying o pulsing, also switching an control functions.
22	DPDT (2-C)	2 amp.	.04 watt DC .02 watt DC	— 2C contacts — 1C contacts	12,000 ohms	1.0 watts 100°C	.004020 sec.			22RJ -	Miniature, double-poli maximum shock resistant and precision.



Nominal, 24100 of 1.1.

life, eg. 100,000 operations.

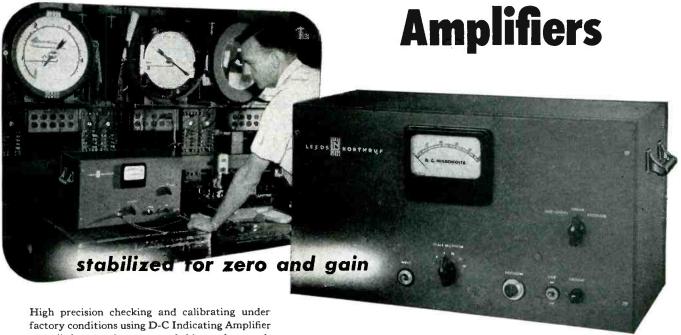
Approximate, as governed by normal variations in ‡ At operate voltage, operated. Unoperated, approx. 1/2 this value. Varies inversely with applied voltage.

§ Sum of ambient plus temperature rise not to exceed temperature given for each relay. Thermal time constant varies from 10 to 30 minutes. For higher temperatures, inquire of factory

Write for fully descriptive catalog

SIGMA INSTRUMENTS, INC.. 62 PEARL STREET, SOUTH BRAINTREE 85, MASS.

For low-level d-c measurements use these new, L&N triple-purpose



as null detector for accurate bridge and potentiometer circuits. Through combination of a-c amplification and balanced feedback network, zero and gain stability are designed right into instrument. Trimmer controls are eliminated.

SPECIFICATIONS

MICROVOLT UNIT No. 9835

MICRO-MICRO-AMPERE UNIT No. 9836

FULL SCALE RANGES WITH BUILT-IN 4" METERS 0 to 50 or -25 to +25 Microvolts; scale multipliers: 1, 2, 4, 10, 20, 40

0 to 1000 or -500 to +500 Micro - Microamps; scale multipliers: 1, 2, 5, 10, 20, 50, 100, 200, 500, 1000, 2000

ACCURACY

Of amplifier: ±0.4% of reading; Of meter: ±1%

Of amplifier: ± 0.5 to 0.8%* of reading; Of meter: ±1%

ZERO OFFSET

Max. offset: ±0.5 Microvolt

Max. offset: $\pm 2\%$ of scale

*SOURCE RESISTANCE

Up to 10,000 ohms.

0.1 megohm or more.

REPONSE TIME

2 to 3* sec.

2 to 3* sec.

OUTPUT

For full scale input on any range: 10 millivolts at output impedance of 500 ohms for null recorder; 1 volt for 20,000-ohm external meter.

Front panel fits standard 19" relay rack.

*Accuracy and Response Time depend on Source Resistance.

can be used as ...

- DIRECT-READING MICROVOLTMETER OR MICRO-MICROAMMETER
- √ RECORDER PREAMPLIFIER
- √ NULL DETECTOR

THESE new D-C Indicating Amplifiers are the answer for all your low-level measurements with thermocouples, strain gages, bolometers—bridge and potentiometer circuits ionization, leakage, and phototube currents-almost any measurement of extremely small direct current or voltage.

Actually 3 instruments in 1, Amplifiers can be used as:

√ Direct-reading instruments—Scale multiplier knob lets you select the range in which you want to work.

√ Recorder preamplifiers—with broad flexibility. One or two degrees temperature difference can be spread right across a 10" Speedomax recorder chart.

√ Null detectors—more sensitive than most reflecting galvanometers, yet with full scale response time of only 2 to 3 seconds. Leveling is unnecessary; the instrument is not affected by vibration. At the turn of a range knob, you have available a wide choice of sensitivities. And when using non-linear response, not only does the instrument stay on scale at extreme unbalance; sensitivity increases automatically as the null point is approached.

For details, send for Folder EM9-51(1). Write our nearest office, or 4979 Stenton Ave., Phila. 44, Pa.

NORTHRUP

Irl. Ad EM9-51(2)



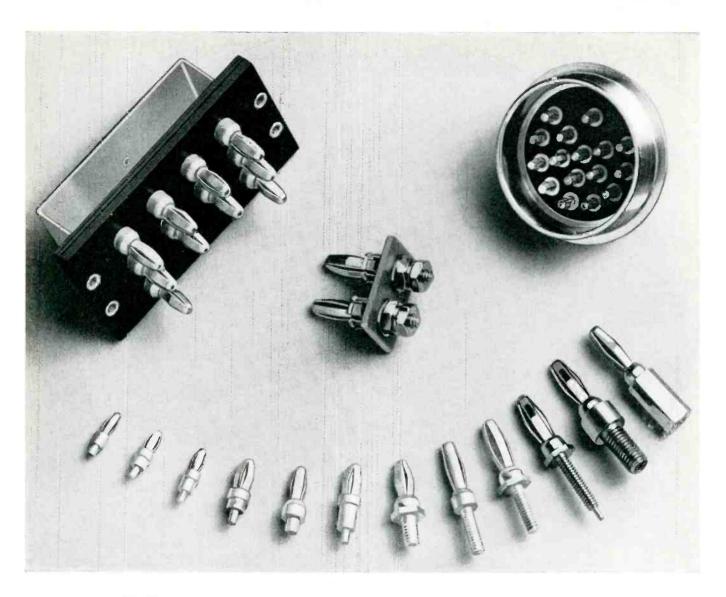
The resulting popularity and industry acceptance of our toroidal coils and filters have necessitated an expansion of our production facilities to ten times what they were five years ago and we are proud to point to this growth as an expression from our customers of their satisfaction in the quality of our product and our service.

EXCLUSIVE MANUFACTURERS OF COMMUNICATIONS NETWORK COMPONENTS

Burnell & Company
YONKERS 2, NEW YORK
CABLE ABORESS "BURNELL"

TC-4 - Ind. Up to 10 Hys.

TC-5 - Ind. MHY-Up to 750



Volume and Variety

You can count on Banana Pins to keep contact in mobile connections, even under conditions of extreme vibration. They come in a wide range of shapes and sizes, and are easily adapted to special assemblies. Here at Ucinite, we have the plant capacity to manufacture standard types in volume and are capable of engineering special types as required. We are also set up to turn out complete banana pin assemblies to your specifications.

Ucinite, with a wealth of experience in the development and production of electric components and assemblies, stands ready as before to supply parts promptly to manufacturers engaged in the present expanding defense production program.

The UCINITE CO.

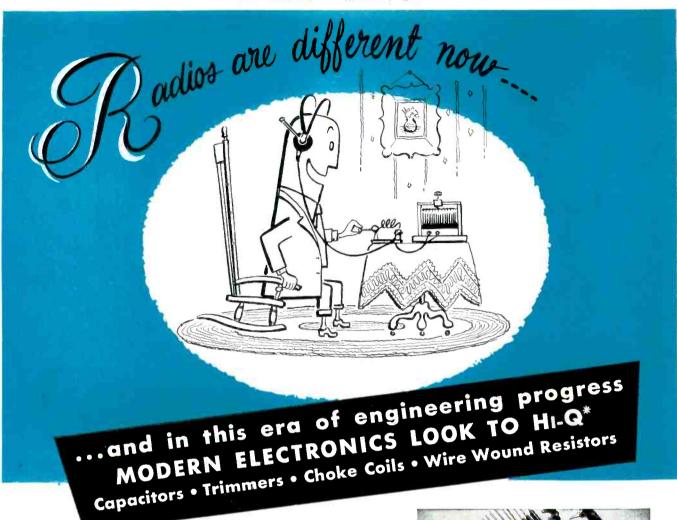
Newtonville 60, Mass.

Division of United-Carr Fastener Corp.

Specialists in

ELECTRICAL ASSEMBLIES,

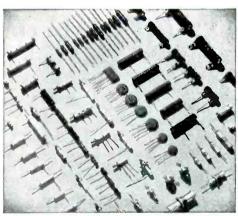
RADIO AND AUTOMOTIVE



• No science has grown faster or gone farther than electronics since those first amazing crystal receivers picked words and music out of the air. And in the development of modern radio, TV, communications and other electronic equipment, one of the prime contributions to dependability, long life and compactness has come from the use of ceramic components as perfected by HI-Q.

Though HI-Q output in four plants is now at the rate of several million units a month, there is no variation in the quality or high performance characteristics of its capacitors, trimmers, choke coils and wire wound resistors. Engineering watchfulness and rigid production control insure the maintenance of uniform precision standards well within the limits of minute tolerances. And it has become an industry maxim that in the development of special components to meet unusual requirements, HI-Q engineers are the most competent and resourceful in their field. Their services are available to your organization in helping solve any problems you may have.





BETTER 4 WAYS

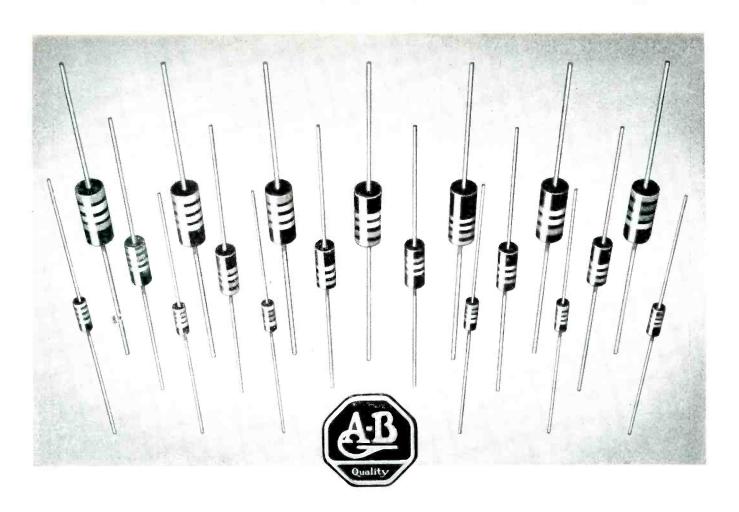
- PRECISION
- UNIFORMITY
- DEPENDABILITY
- **MINIATURIZATION**



*Trade Mark Registered, U.S. Patent Office

Electrical Reactance Corp.

SALES OFFICES: New York, Philadelphia, Detroit, Chicago, Los Angeles PLANTS: Olean, N.Y., Franklinville, N.Y. Jessup, Pa., Myrtle Beach, S. C.



FIXED RESISTORS THAT STAY FIXED ... because they are rated at 700



Bradleyunits have a wide safety factor. They are not crowded for performance because they are rated at 70C . . . not at the usual 40C.

Another thing . . . under continuous full load for 1000 hours, the resistance change is less than 5 per cent. They withstand all extremes of temperature, pressure, and humidity. And, don't forget that Bradleyunits require no wax impregnation to pass salt water immersion tests.

Bradleyunits are packed in honeycomb cartons to keep the leads straight and to avoid tangling during assembly operations. They are available in $\frac{1}{2}$, 1, and 2-watt ratings in standard R.T.M.A. values up to 22 megohms.

Allen-Bradley Co.
110 W. Greenfield Ave., Milwaukee 4, Wis.



FOR EXTRA FAST

STOPS



REVERSES

Holtzer-Cabot Type R-24 motors

are designed especially to provide the rapid starts, stops and reverses required by modern, high-speed process control equipment. Extremely low rotor inertia, plus high accelerating and decelerating torques, provide extremely quick response. Tests show that, with the dynamic braking connection, the unloaded rotor will always stop within three-quarters of one revolution! On geared motors, of course, this means that the output shaft stops in a very small fraction of a revolution.

These motors are 2-phase, squirrel-cage, nonsynchronous induction type. For operation on a single-phase circuit, a capacitor is used in series with one phase. Special high impedance windings are available, suitable for matching standard power tubes.

Available speeds, 1300 RPM (1800 synchronous speed) to 3/4 RPM (geared), with torque ratings from 0.5 to 75 oz. in. Can be provided for odd voltages and frequencies, with ball or sleeve bearings. These motors are ideal for Servo-Mechanisms.



INVESTIGATE NOW... Holtzer-Cabot welcomes inquiries involving special motors

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For information on the availability of Alnico Permanent Magnets, write today. Carboloy's engineering consultants and technicians will give immediate attention to your magnet problems.

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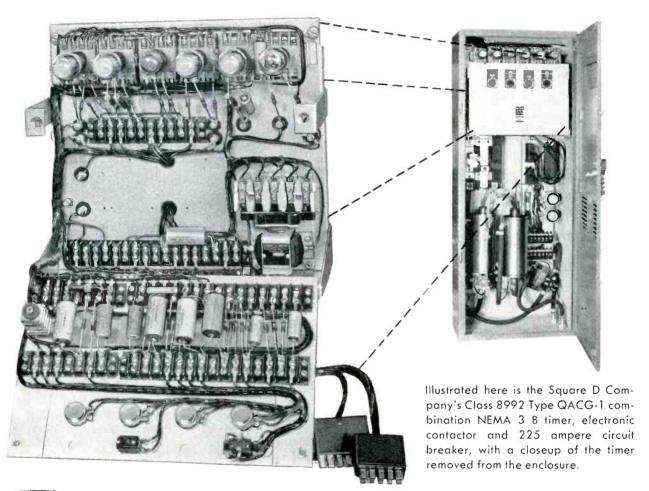
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ELECTRONICS — July, 1951

WIRING KNOW-HOW CONTINUING PROOF . . .



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Meticulous workmanship is vital in the UNILEC-TRIC Wiring System installed in this timer for the Square D Company's Electronic Welder Control. In addition, this application has resulted in reported savings up to 20% for the Square D Company, plus savings in installation time and inspection.

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July, 1951 — ELECTRONICS



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But we never realized just how popular it would be. This inexpensive turntable tape reproducer has turned into the sensation of the industry—

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- No motor-turntable acts as motor.
- Can be plugged into any standard speech input equipment.
- Speeds of $7\frac{1}{2}$ " and 15" per second, response up to 15,000 cps.
- Low cost—but has reproduction quality and fidelity of a high priced machine.
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Outstanding design and precision workmanship assure completely pressurized electrical connectors for all sizes of contacts. A truly important feature, but only one of the many exclusive advantages that contribute toward making Bendix outstanding in the electrical connector field. Increased resistance to flash over and creepage is made possible by the use of Scinflex dielectric material—an exclusive development of Bendix. In temperature extremes, from -67° F. to $+275^{\circ}$ F. performance is remarkable. Dielectric strength is never less than 300 volts per mil. Remember, for the greatest value in electrical connectors, it pays to specify Bendix. Our sales department will gladly furnish complete information on request.







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CONTROLS... A Two-Edged Sword

It may seem dangerously premature to talk about getting rid of emergency government controls while all-out war is still an imminent possibility.

But success in this strange struggle for our freedom into which the Russian Communists have plunged us requires that we:

- 1. Maintain a whole battery of controls designed to speed defense production and curb inflation, and at the same time
- 2. Work to end the controls at the earliest possible moment.

Here is the reason why this editorial—fourth in a special series on mobilization for defense—is devoted to the need for a speedy release from controls.

If the Russian Communists can force us to maintain indefinitely the present system of government controls, they will have won a tremendous victory. They will have saddled us with a system of collectivism which, over a period of years, would be fairly certain death to freedom of business enterprise.

Make no mistake about it. This is not an argument against emergency controls. We need controls now to break a right of way for

our mobilization program through the business boom. Indeed, the third editorial in this special series was titled "Why Controls Are Necessary." It stressed both the need for controls and the need for positive cooperation to make them work.

Controls Can Undermine Our Economy

But these controls surely chisel at the foundation of our normal economic system. So long as we have them, many if not most key business decisions will be made in Washington bureaus rather than in the free market place. For example, the *National Production Authority* administers a *Controlled Materials* Plan (italics ours) which directs the flow of basic metals, and decides who can use them for what purposes.

Happily, the people who operate these controls are not using the methods of a secret police state.

Even more happily, most of the leaders who have been drafted to manage the controls are not in love with their jobs. They are doing their best in the thankless task of making controls work. They recognize the danger of chronic controls.

But the fact remains that our economy is

operating under arrangements which carry it a long way toward the pattern of centralized control the Russians would inflict on the world.

The Wilson Plan

A plan for getting rid of these controls has already been developed. It was put together by our Director of Mobilization, Charles E. Wilson—while he was working day and night to set up the necessary emergency controls.

The Wilson Plan—if we escape all-out war—will strengthen our defenses and our economy. By 1953, it calls for:

- 1. Providing the weapons to equip an armed force of 3½ to 4 million, together with a supply of weapons for our allies.
- 2. Building a stockpile of weapons which, with current production, would be sufficient to carry on an all-out war for a year.
- 3. Building the manufacturing capacity by which we could rapidly expand our production of weapons if all-out war should come.
- 4. Increasing the productive capacity of industry enough to resume the expansion of our civilian economy.

With these jobs done our economy would be big enough and strong enough to meet both civilian and military requirements. And the government controls needed for mobilization could be speedily dropped.

Call for Sacrifice

The Wilson Plan requires a major effort—it means spending more than \$50 billion a year for mobilization. That is almost 20 percent of our total production. And this cannot be done without sacrifice. For a time, particularly in the next year, living standards will drop.

But the sacrifice required is amazingly small. At the peak of the defense effort, civilians will still have available to meet their needs about as much as they did in any year before 1948.

To make the Wilson Plan succeed we must curb inflation. A second year of inflation such as that which we have had since the Korean war started would multiply disastrously the costs of our defense program. One key part of a successful program to curb inflationary pressure, which soon will be building up again, is a pay-as-we-go tax program. The second editorial in this series urged that we do our utmost to pay as we go.

We Cannot Out-Control the Communists

But, above all, to make the Wilson Plan work we must keep our sights set on the crucial importance of increased production. Our problem is to increase our capacity to produce so that we can carry both a major military program and an expanding civilian economy for as many years—General Bradley thinks it might be fifteen or twenty—as the menace of Russian Communist aggression persists.

If we do not produce enough to do this double job, we shall be confronted with the prospect of having to live indefinitely under government controls of the sort that have been set up since the start of the Korean war. That would be delightful to the Russian Communists. It would go far toward making over our economy on the Moscow model.

Even if we wanted to, we never could hope to out-control the Russians. They are miles ahead of us in that line. But we can out-produce them, by a tremendous margin. By doing that we shall travel the surest road to victory.

McGraw-Hill Publishing Company, Inc.

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These AlSiMag spacers are flat within .0002" and parallel within .0005" Thickness tolerance \pm .001". For special applications, flat shapes have been ground to a thickness of .008".

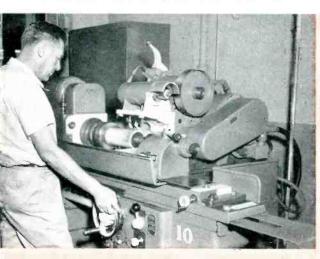
- We have advanced the technique of grinding AlSiMag ceramics so that we can obtain, on critical dimensions which lend themselves to grinding, the accuracy of comparable precision metal parts.
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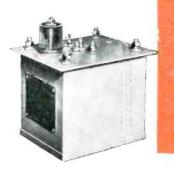
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Resonant-charging reactors, accurately designed and constructed for radar service. Usually required in ratings of 40 kv and below, 1 ampere and below and 300 henries and below. Higher ratings are being built, and can be considered. When required, small- and medium-size designs can be provided with 3 to 1 range of inductance adjustment.

Filament Transformers

Filament transformers available with or without tube socket mounted integral with the high-voltage terminal. Low capacitance. Ratings to match any tubes; insulated to practically any required level.



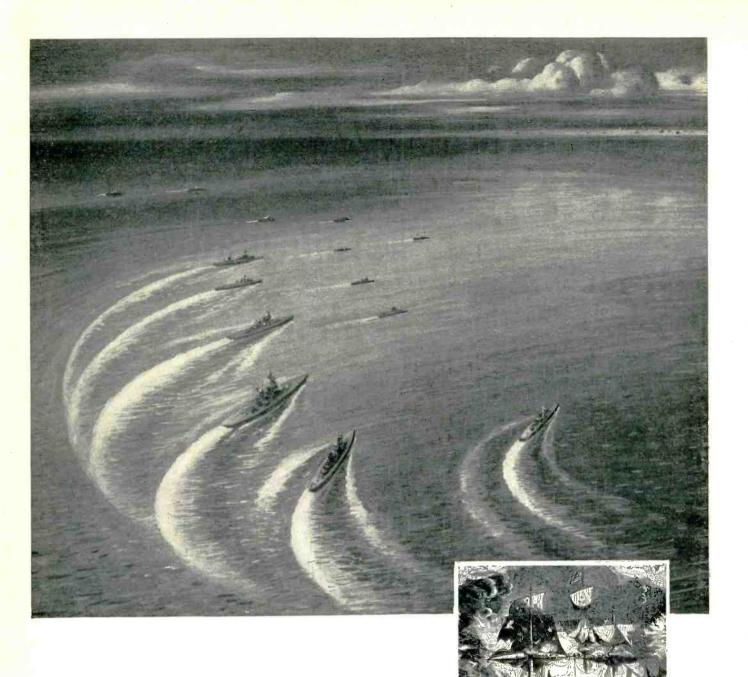




Illustrated here are typical high-voltage components manufactured by General Electric. They can be built to meet Armed Services requirements. All are oil-filled and hermetically sealed—with excellent ability to withstand mechanical shocks and to operate continuously for long periods in widely varying temperatures. Apparatus Dept., General Electric Company, Schenectady, N. Y.

Your inquiries will receive prompt attention. Since these components are usually tailored to individual jobs, please include with your inquiry, functional requirements and any physical limitations. Write to Apparatus Dept., 42-328A, General Electric Co., Pittsfield, Mass.





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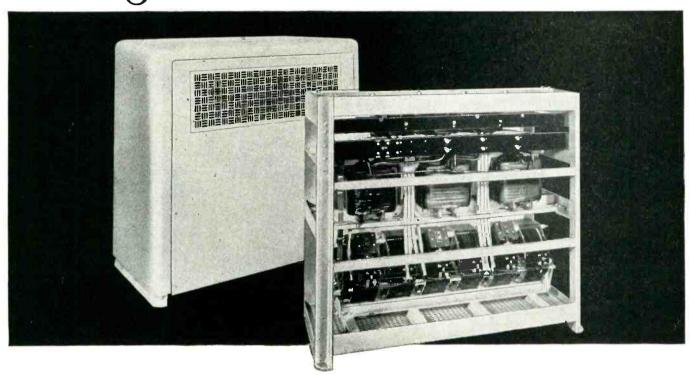
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The Westinghouse solution: A combination of a three-phase, dry-type transformer to produce the required 4,000 volts, with a three-phase, buck-boost transformer, and a three-phase, motor-driven power-stat, serving a d-c rectifier. Quality-proven class B insulation and HIPERSIL® cores result in compact, lightweight installation that completely eliminates the need for bulky, oil-immersed, tap-changing equipment.

The main transformer is rated 73 kva, 230 volts delta on the primary, 4,240 volts wye on the secondary. The secondary of the buck-boost is in series with the output of the main transformer, and the powerstat acts to adjust the primary voltage on the buck-boost, so that stepless control of the entire assembly can be achieved under full load of the d-c rectifier. The operator stands in front of the transmitter itself, and raises or lowers the voltage by means of pushbuttons, while he watches the d-c plate voltmeter of the rectifier.

If you have a tough transformer problem, take advantage of the facilities of Westinghouse for quick, practical solutions. Transformers specially designed for all types of electrical and electronic circuits, as well as a wide selection of standardized designs . . . produced in quantity . . . with quality. Call your nearby Westinghouse representative, or write Westinghouse Electric Corp., P. O. Box 868, Pittsburgh 30, Pa. J-70565

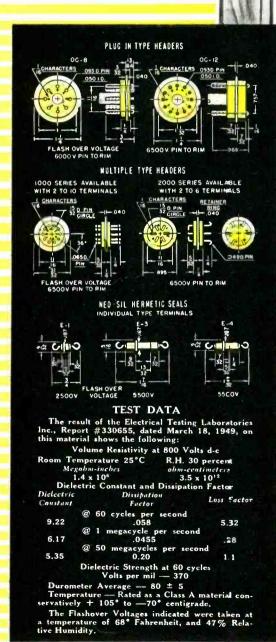




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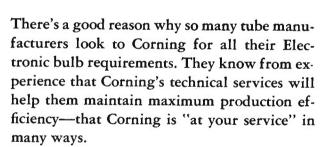


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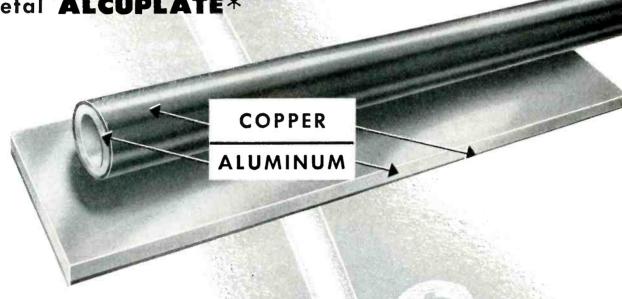
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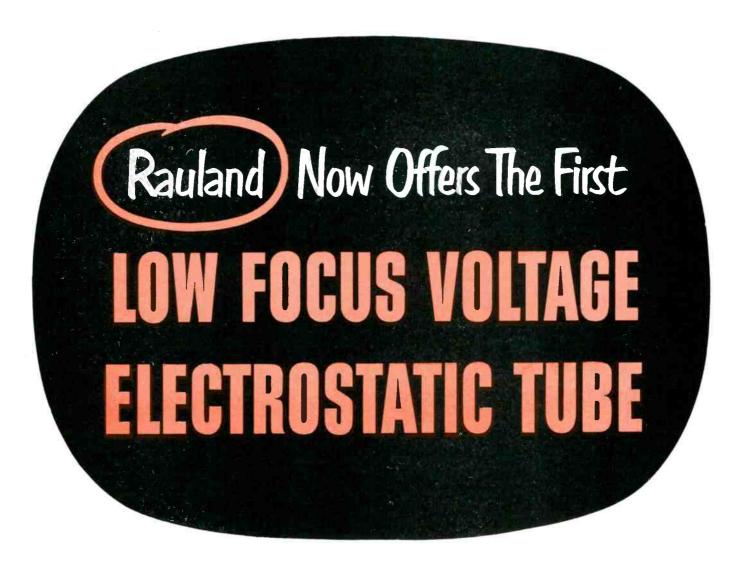
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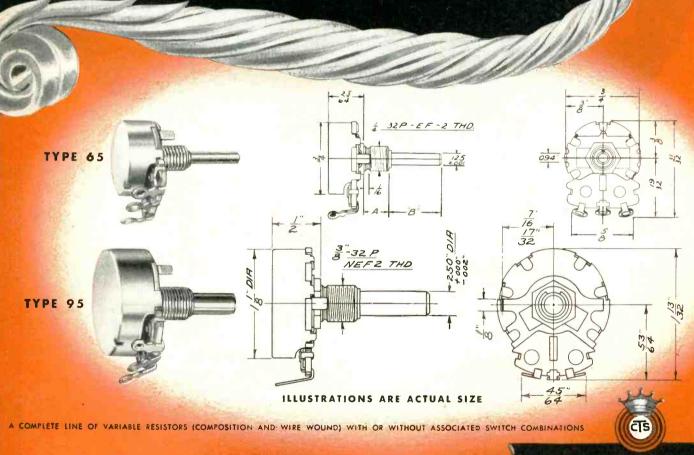
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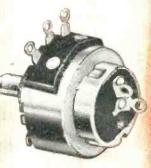
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JAN Type RA 20A 2 Watt (CTS Type 252)



JAN Type RA 20B 2 Watt (CTS Type GC-252)



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RA 25	5A	25	(without switch)	3
RA 25	5B (3C-25	(with switch)	3
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Uses

Time bases, rate indicators, clock systems, chronographs, geo-physical prospecting, control devices and for running small synchronous motors.

Jeatures

- 1. Bimetallic, temperature-compensated fork, no heating or heat-up time is required.
- 2. Fork is hermetically sealed, no barometric effects on frequency.
- 3. Precision type, non-ageing, low coefficient resistors used where advantageous.
- 4. Non-linear negative feedback for constant amplitude control.
- 5. No multi-vibrators used.
- 6. Synchronous clock simplifies checking with time signal.

Specifications

Accuracy—1 part in 100,000 (.001%). Temperature coefficient-1 part in 1,000,000 per degree centigrade (or better). Outputs-

- 1. 60 cycles, sine wave, 0.110 volts at 0 to 10 watts (adjustable).
- 2. 120 cycle pulses, 30 volts negative.
- 3. 240 cycle pulses, 30 volts positive and negative. Pulse duration, 100 micro-seconds.

product of

AMERICAN TIME PRODUCTS

580 Fifth Avenue

New York 19, N. Y.

Operating under patents of the Western Electric Company



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AND GET STRONGER, TIGHTER FASTENINGS, TOO!

separate tapping is completely eliminated no tapping time or tool costs!

each screw fits tight in its own self-cut, mating thread—powerfully resists loosening!

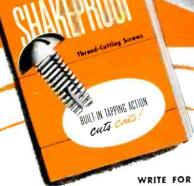
greater strength permits use of smaller or fewer fastenings—saving space and weight.

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with spaced threads
to minimize stripping

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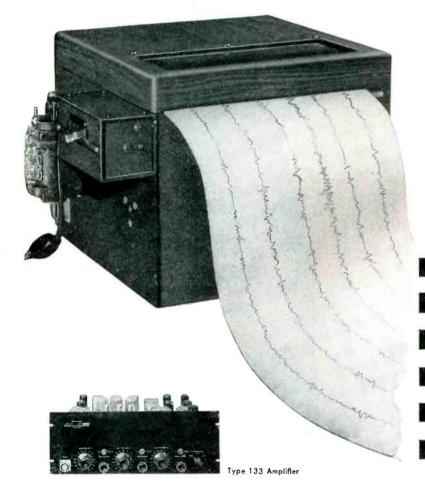
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THE OFFNER DYNOGRAPH RECORDER



drift-free d-c recording

SPEED

TEMPERATURE

PRESSURE

VIBRATIONS

POSITION

OTHER VARIABLES

check these exclusive Dynograph features

Pen Deflection Linearity of 1% with pen response of 1/120th of a second.

Sensitivity of 150 microvolts d-c per centimeter of pen deflection.

Stability and drift-free operation through a special chopper type amplifier.

No extra equipment needed with reluctance type pick-ups.

True Differential input obtained through special transformer coupling.

Offner Electronics Inc., the original manufacturer of precision, direct writing oscillographs, now offers to industry the Dynograph Recorder—a direct writing, high speed oscillograph with microvolt d-c sensitivity.

Operating at approximately 100 times the speed of other industrial recorders with comparative sensitivity, the Dynograph simultaneously records many variables formerly recorded only by photographic means.

Now you can obtain a precise record of transients in the operation of various equipment through a ruggedly built, easy to maintain, versatile d-c recorder—the Dynograph.

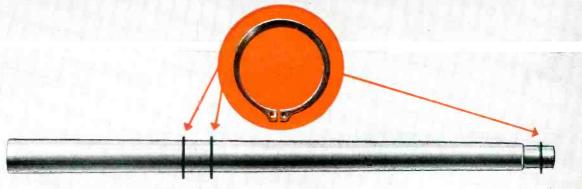
Write Today for Bulletin L-311 — Complete
Specifications and Construction Details of the Dynograph

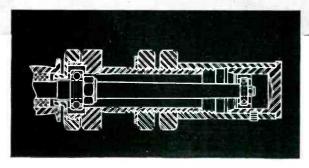
OFFNER ELECTRONICS INC.

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CHICAGO, 25

3 TRUARC RETAINING RINGS LOWER COST... IMPROVE PERFORMANCE OF REVOLUTIONARY NEW TEXTILE SPINDLE!

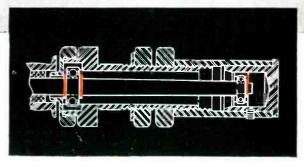




OLD CONSTRUCTION. To position 2 ball bearings, an oversize diameter rod had to be turned on a lathe to provide 3 shoulders. In addition, blade required 2 threading operations . . . 2 lock nuts . . . separate tapering operation. Proper pressure of nuts against ball bearings required skilled labor adjustment.

The H & B American Machine Company's new antifriction CENTURY spindle is probably the most mechanically advanced spindle on the market today. Waldes Truarc Retaining Rings have eliminated many of the material, tooling and assembly costs...have kept its price competitive. Truarc Rings not only simplify spindle assembly, they position ball bearings accurately...simplify maintenance...eliminate skilled labor...improve performance! And there are Truarc Rings to solve any design or re-design problem!

Redesign with Truarc Rings and you too will cut costs.



NEW CONSTRUCTION. Standard rod, equal in diameter to finished blade is used. Three grooves for Truarc Rings and shoulder are made quickly and easily on screw machine. Blade is economically tapered by centerless grinding. Truarc Rings maintain correct pressure on ball bearings for life of unit!

Wherever you use machined shoulders, bolts, snap rings, cotter pins, there's a Waldes Truarc Retaining Ring designed to do a better job of holding parts together.

Truarc Rings are precision-engineered...quick and easy to assemble and disassemble. Always circular to give a never-failing grip. They can be used over and over again.

Find out what Truarc Rings can do for you. Send your blueprints to Waldes Truarc engineers for individual attention, without obligation.

Waldes-Truarc Retaining Rings are available for immediate delivery from stock, from leading ball bearing distributors throughout the country.

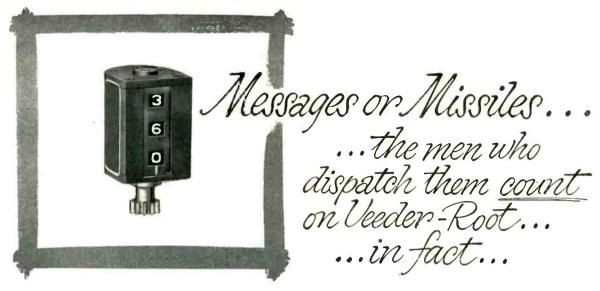


RETAINING RINGS

WALDES KOHINOOR, INC., LONG ISLAND CITY 1, NEW YORK WALDES TRUARC RETAINING RINGS ARE PROTECTED BY THE FOLLOWING PATENT NUMBERS: 2.382,948; 2.416.852; 2.420,921; 2.428,381; 2.487,802; 2.487,803; 2.491,306; 2.491,310; 2.509,081; AND OTHER PATS. PEND.

0	Waldes Kohinoor, Inc., 47-16 Austel Place	E073
H	Long Island City 1, N. Y.	
	Please send Bulletins 6, 7 and 8—giving engineerin specifications for all types of Waldes Truarc Ring	-
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ELECTRONICS — July, 1951



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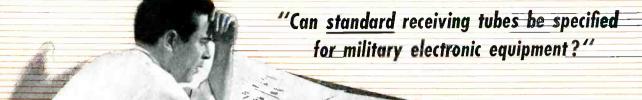
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In some cases, no. But all General Electric 5-STAR TUBES are on the Armed Services' Preferred List. They're the last word in dependability! Every 5-Star Tube gets 48 hours' service at the factory under Class A conditions. Frequent life tests also are made. During manufacture, 5-Star Tube parts are inspected individually prior to careful and precise hand-assembly.

That new military circuit you've been asked to design, will do its task better-far more reliably-with G-E 5-Star Tubes to help assure steady performance under fire. You have nine types to choose from, covering a wide range of applications. Other types will be added later. They will further extend G-E 5-Star usefulness.

Aviation—for which 5-Star Tubes originally were developed pays enthusiastic tribute to the job these special G-E tubes have done in making airborne travel safer. Now General Electric offers the same fine tube performance to you, so that equipment you're designing may approach absolute standards of reliability- may meet stiff military performance requirements. Fill out, clip, and mail the coupon below! Electronics Department, General Electric Company, Schenectady 5, New York.





Sharp-cutoff r-f pentode H-f twin triode Beam power amplifier Dual-control sharp-cutoff r-f pentode Twin diode Remote-cutoff r-f pentode Pentagrid converter. High-mu twin triode Medium-mu twin triode

General Electric Company-Section 4, Building 267-Schenectady 5, N.Y.

Please mail me the facts about G-E 5-Star Tubes—their performance characteristics; how they are manufactured, how tested.

Please have a tube engineer call on me, to discuss G-E 5-Star Tubes and their applications in circuits I am designing.

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GENERAL 🟀 ELECTRIC



When performance of the promise!

We promised you better resistors than had been available before. Our new crazeless, blue-gray enamel resistors are out-performing our highest expectations in the field.

Let us summarize the improvements you get in these new Hardwick, Hindle resistors:

- 1. Crazeless—complete elimination of the crazing which results in failure of the resistive element due to moisture penetration from humidity, salt and severe atmospheric conditions—providing greater dielectric strength.
- 2. High temperature qualifications provide greater safety factor under abnormal operating conditions.
- 3. Stronger core—gives greater resistance to excess vibration and shock test.
- 4. Alloy terminals—greater strength—corrosion and rust resistant—spot-welded to the ceramic body.
- 5. Resistive element—finest quality wire free from impurities that could cause weakness.

6. Connections—all wire connections to terminals are protected by positive non-corrosive bonding.

The above factors give you <u>absolute</u> <u>assurance</u> of longer life and outstanding performance.

The fixed, the ferrule and the flat types are designed for and manufactured in accordance with JAN-R-26A specifications.

HARDWICK, HINDLE, INC.

Rheostats and Resistors

Subsidiary of

THE NATIONAL LOCK WASHER COMPANY

Established 1886

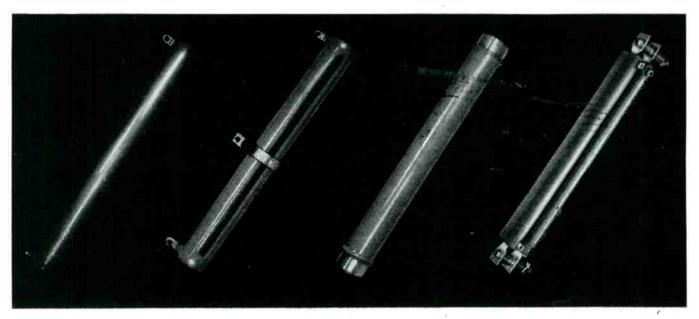
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PICTURE this trade mark in your mind when you want cabinets and housings to enhance your finished products with step-ahead styling...absolute uniformity...ease and economy in your assembly operations... greater service benefits to the user. The Blueprint Man symbolizes all this, on jobs both simple and exacting... at prices figured with a sharp pencil. Write for our data book.

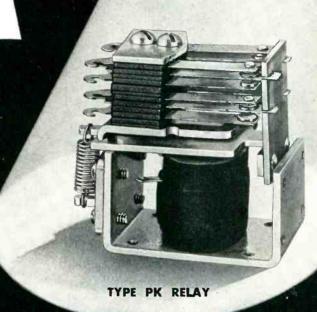
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HERE ARE THE FACTS AND FIGURES:

CONTACTS: 10 amp. standard. 24 volts D.C., 115 volts A.C.

15 amp. contacts available.

SENSITIVITY: D.C.: 4 pole 1.5 watts 2 pole .7 watts

A.C.: 4 pole 5 volt amperes 2 pole 2.5 volt amperes

Can also be furnished in 6 pole AC and DC up to 4000 Ohms.

COIL: To 115 volts D.C., 230 volts A.C.

NOMINAL HEAT RISE: D.C. 30°C above room ambient A.C. 45°C above room ambient

MAX. INPUT FOR 85° RISE: D.C. 5 watts A.C. 11 volt amperes

MOUNTING: Base or end mounting

WEIGHT: 4.5 oz. 4 P.D.T.

WEIGHT HERMETICALLY SEALED: 7.7 oz.

DIMENSIONS: Open Relay—21/16", 11/8", 21/16"

Sealed Relay-31/8", 11/2", 25/16"

Overall Mounting Flange—31/8"
Center to Center Mounting Holes—211/16"

A Quality Relay

The new Allied PK Relay is designed to offer versatility in a power relay where quality and ow cost are factors. Besides stability in operation its reliability allows a range in applications from high quality instruments to vending machines. The PKU relay will comply with Underwriters' Laboratories requirements and can also be supplied hermetically sealed.

Bulletin PK gives complete details. Send for your copy roday.

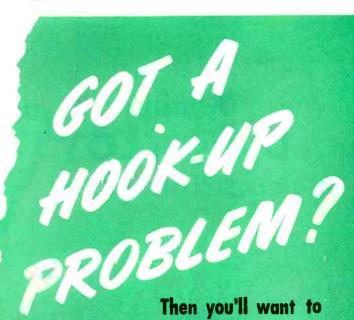
Be sure to send for your copy of Allied's Relay Guide. It gives the engineering data for 27 Allied telays in a concise tabular form for easy reference.



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ALLIED CONTROL COMPANY, INC. 2 EAST END AVENUE, NEW YORK 21, N.Y.





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Well, right off the bat, there's the fact that many types of Synkote Other Synkote Hook Up Wire are approved constructions, conforming to JAN-C-74

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Or you can pick other constructions for special requirements. industry is working on today. If yours is a "hot spot" application, for instance, use Synkote wire that carries UL approval for continuous operation at 105°C.

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Just tell us your requirements in electronic wiring. We'll be glad to help you select, from the comprehensive Synkote line, the type of insulated conductor that best meets your needs. And we're prepared, on short notice, to engineer new designs to fit your problem — and to produce them in practically any quantity.

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PLASTOID CORPORATION, Dept. E 42-61 24th Street . Long Island City 1, N. Y.

Please send me additional information on Synkote Hook Up

Wire Type.

Name Company Address

State



Once more, RCA leads the field! This time, with new rectangular picture tubes which require no focusing coil or focusing magnet . . . thus effecting important savings in critical materials.

The currently used magnetic focus in TV sets requires coils and magnets containing the metals cobalt and copper... both on the Government's restricted list.

To meet the shortages, RCA engineers have developed improved electrostatic focusing for wide-deflection-angle TV tubes—eliminating need for a focusing magnet or coil. Result: The new RCA kinescopes—which not only make possible savings in critical materials, but introduce advantages that destine them to become industry's most widely used picture tubes.

Such advantages, for example, as: (a) An improved electron gun which provides excellent uniformity of focus over the entire picture area, and is so designed that the focusing electrode takes negligible current—permitting voltage for the electrode to be supplied easily and

economically; (b) Focus automatically maintained with variation in line voltage and adjustment of picture brightness; (c) Simplification of tube installation and adjustment for optimum performance.

When RCA produced its outstanding Image Orthicon, "eye" of the television camera, it called upon the superlative Driver-Harris radio alloy Nichrome* V to provide 95% of the metal components of the tube. Now, in the case of the kinescopes, a Driver-Harris produced alloy fills a need.

Here are typical examples of how Driver-Harris stands ready to serve all industry with alloys necessary for new or standard applications.

During the present emergency, of course, strategic materials and the alloys made from them are on strict allocation. However, we shall be glad to make recommendations based upon your specific needs, and serve you to the best of our ability.

Makers of world-famous Nichrome* and over 80 other alloys for the electronic, electrical and heat-treating fields

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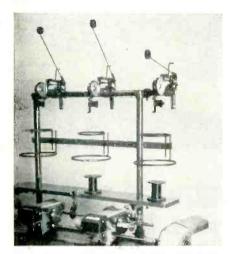
Denvert Kansas City, Mo. Newark Pittsburgh San Francisco Albanyt Chicago New Drieans Atlanta Cincinnati Detroit Los Angeles Providence Seattle Cleveland Waterbury Baltimore Houstont Milwaukes (tsales office only) Dallas Indianapolis Minneapolis Philadelphia St. Louis



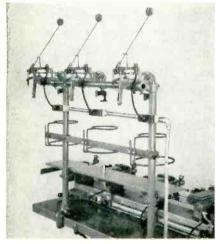
FREE Chase Book lists mesh, diameter of wire, per cent of open area, weight and other important data.

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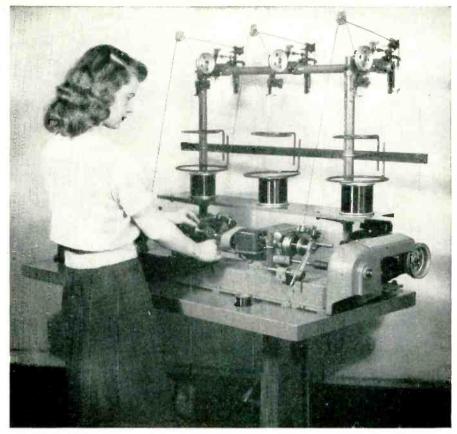
NON-INSULATED COILS WOUND AT 5000 RPM



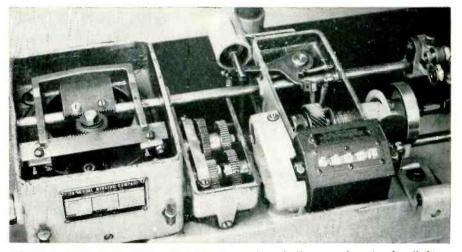
WINDING STARTS INSTANTLY WITHOUT DAN-GER OF WIRE BREAKAGE with the overend tension. The first turn can be started tightly against the spool or bobbin head, improving "lay".



WIRE BREAKAGE DETECTOR is optional equipment for stopping arbor promptly when wire breaks or runs out. This relieves operator of having to watch wire spool continually and prevents counting of turns when wire is not being wound.



WHEN CPERATOR FINISHES manual procedure on one head, another coil will have just been completed. By synchronizing winding and handling time on the No. 102 Universal Hi-Speed Coil Winder, you get maximum production per operator. Coils up to 15,000 turns can as a rule be handled efficiently on a 3-head machine at a maximum of 5000 rpm.



READILY-ADJUSTABLE TRAVERSE MECHANISM permits winding any length of coil from 1/16 in, to 2 13/16 in, without changing cams. A single setting applies to all the winding heads.

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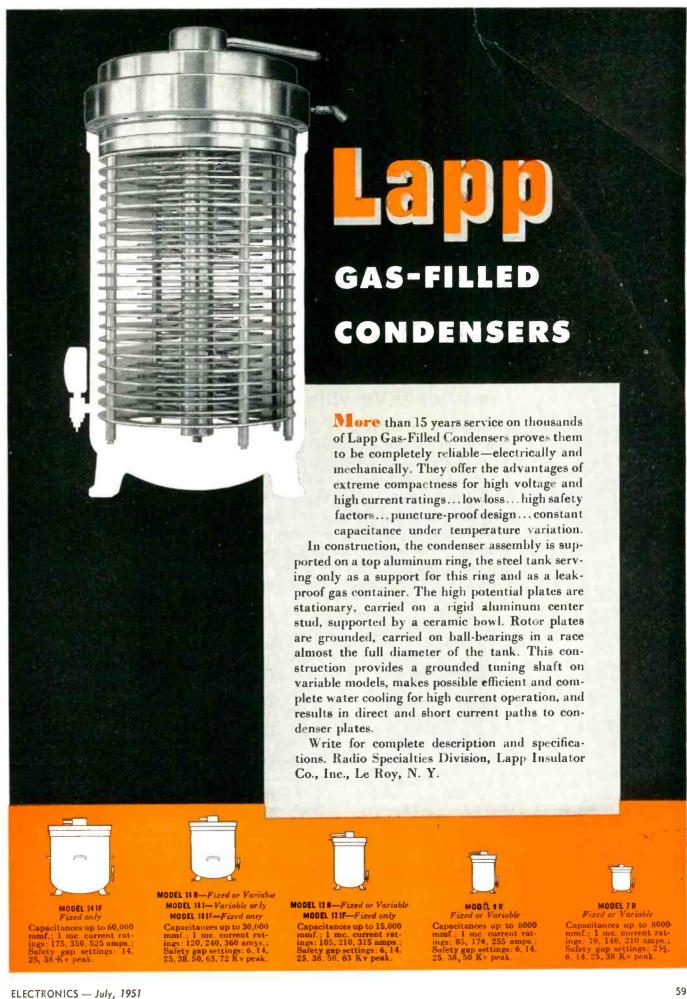
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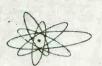


For winding coils in quantity accurately . . . automatically use Universal Winding Machines

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ELECTRONICS



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HIGH VOLTAGE COMPONENTS

- Withstand Mechanical Shocks
- Operate Continuously for Long Periods in Widely Varying Temperatures

Available in a wide range of ratings, high-voltage components shown here are typical of units manufactured by General Electric for applications 5000 volts and above where corona must be held to a minimum. They represent many years of experience in meeting Armed Services requirements, and can be built for today's military specifications. Because these components are usually tailored for each job, please include functional requirements and any physical limitations with your inquiry. Write to 43-328A, General Electric Company, Pittsfield, Mass.



RECTIFIERS — A-c to d-c powersupply units especially built for precision work where unusually low regulation, light weight, and small size are necessary. Typical outputs available are 7, 9, and 13 kv. Illustrated 7-kv unit measures 6 x 6 x 7 in., weighs 8 lb.



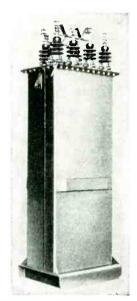
ruse with either hard-tube or line-type modulators. Available in peak-voltage ratings from 10 to 100 kv or higher; peak-power ratings up to 30,000 kva or more. Designed for operation on pulse durations from 0.1 to 20 microseconds at rates up to 4000 pps.



rately designed and built for radar service. Currently available in peak operating voltages from 5 to 45 kv, current ratings up to 2.25 amp. Higher ratings can be provided. Inductance ratings ranging from 0.25 to 300 henrys remain constant within 5 percent at above 50 percent load current.



For special applications or for use with standard high-voltage rectifier tubes. Supplied with or without tube socket mounted integrally with high-voltage terminal. Insulated to nearly any required level.



MODULATION TRANSFORMERS

- High-fidelity, low-phase shift, Pyranol* or oil-modulation transformers and reactors for high-power AM transmitters are available as integral or separate units. Highly developed designs in ratings from 3.5 to 500 kw provide wide transmission frequency range, keep down harmonic-voltage insertion. Low phase shift gives flexibility and range.

*Reg. TM of General Electric Co.

GENERAL



ELECTRIC

667-16

TIMELY HIGHLIGHTS ON G-E COMPONENTS



Signals with band widths up to 2 megacycles can be delayed from 0.25 to 10 microseconds. Available in lengths up to 100 ft. Delay equals approximately 1/2 microsecond per foot. Characteristic impedances of 1100 and 400 ohms per foot are available. By ordering in bulk, lengths can be cut to fit specific needs. Can be bent into 4-inch diameter coils. Uses include research and development of special circuits for electronic devices. See Bulletin GEC-459 for further information.



For accurately detecting, controlling, and measuring light and for detecting and measuring even the smallest variations in colors. These cells are especially useful where long life and stability are required or where electronic amplifiers are not practical. Available in a new hermetically sealed series with standard mounting and a wide variety of unmounted sizes. More G-E photovoltaic cells than any other make are used in scientific instruments. Characteristics, dimensions, circuits, and technical data are available in Bulletin GEC-690.



TRANSFER AND CONTROL SWITCH-OVER 10.000 POSSIBLE COMBINATIONS

Built for reliability and long service life, the G-E Type SB-1 transfer and control switch can be used for more than 10,000 possible circuit-sequence combinations. Precision construction permits as many as 40 stages - four banks of ten stages each - to be operated in tandem. Switches with up to 16 stages and 12 positions are commonly furnished. Ratings go up to



20 amp at 600 volts a-c or d-c. Standard components are interchangeable. Complete description in Bulletin GEA-4746.



SELENIUM RECTIFIERS

. HIGH-VOLTAGE UNITS HAVE LIFE EXPECTANCY OF OVER 60,000 HOURS

Now available from General Electric, these 26-volt RMS selenium rectifier cells have a continuous-service life expectancy of over 60,000 hours. Their initial forward resistance is very low and samples tested after 10,000 hours of operation show an average resistance increase of less than 6 per cent.

The high-voltage output means that stacks made up of these units are about 25 percent smaller than is possible with 12-volt cells. Low resistance means cooler operation and the space saving that goes with it.

If your application calls for compact selenium stacks for use in cramped quarters, these cells provide the solution. Stacks made with the new G-E cells may be obtained with rated outputs from 18 to 126 volts d-c at 0.15 to 3.75 amp. Check Bulletin GEA-5280.



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Name				
Company				
Address				
City	State			

ELECTRONICS -- July, 1951 61



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seamanship...

When the going gets rough, the safety of a ship depends largely on the seamanship of its crew and the good judgment of its officers. To help give the master of a ship full data on conditions contributing to the situation at hand, extraordinary new electronic devices ferret out underwater information of great value.

The Edo Corporation has become a leader in the design, development and manufacture of many of these new underwater detection devices which have far greater accuracy, range and dependability than previously thought possible.

Edo sonar equipments are being produced in ever increasing quantities to help give ships of the United States Navy better underwater "eyes" for both navigation and submarine detection.

OVER A QUARTER OF A CENTURY

Edo's emergence as a leader in the field of sonar development is the result of more than a quarter of a century of experience in the marine and aviation field.

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tion for space and weight saving.

If you'd like to learn more about Edo—our background, facilities, and products—we'll be glad to send you a copy of our 25th anniversary booklet. Just drop a note to Dept. M-7, Edo Corporation, College Point, L. I., N.Y.



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DIFFERENT PHOTOTUBES

"CINTEL" PHOTOTUBES

are manufactured with three different types of cathode surface as follows:-

- TYPE 'A' Antimony-Caesium, sensitive to daylight and blue light
- TYPE 'B' Bismuth etc. Caesium, sensitivity similar to that of the human eye.
- TYPE 'S' Silver-Oxygen-Caesium, sensitive to red-infra-red light.

There are over seventy different types covering every known application, and included in this

* Push-pull types for double sound tracks * Special cells for dye image sound tracks * Cells of high insulation, linearity and stability for accurate photometric work * Cells for use in the ultra violet region of the spectrum * Multiplier cells * Transparent cathode cells, etc.



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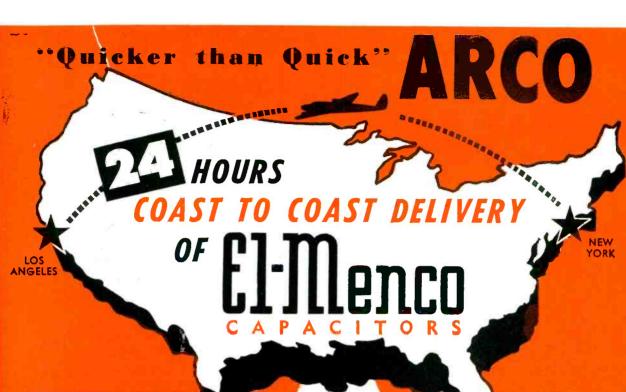
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TYPE VTA FLAT TRANSPARENT CATHODE

TYPE VS. 39 FOR A C C U R A T E PHOTOMETRIC







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CM25	Foil	ог	Silvered	Mica	Cap.	3 MMF-	2.000	MME
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Also A Complete Line Of:-

Trimmers Ranging From 1.5 MMF to 780 MMF Padders Ranging From 15 MMF to 3,055 MMF Paper (Steatite) Ranging From .001 MFD to .5 MFD Ceramic (Silver High "K" Capacitors) From 270 MMFD to 12,000 MMFD

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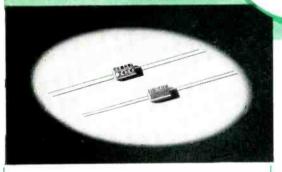
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That's what it cost a prominent watchmaker to develop a type of oil suitable for its precision requirements. (Each worker is given a drop a day with which to oil 500 new watches.)



CM-15 MINIATURE CAPACITOR

Actual Size 9/32" x 1/2" x 3/16" For Television, Radio and other Electronic Applications.

2 mmf. 420 mmf. cap. at 500v DCw. 2 mmf.-525 mmf. cap. at 300v DCw.

Temp. Co-efficient ± 50 parts per million per degree C for most capacity values.

6-dot color coded.

MOLDED MICA

That quantity is no indication of quality is amply demonstrated by the CM-15 El Menco Capacitor. Tiny as it is, it will give sustained superior performance under the most adverse conditions.

Before it leaves the factory it is tested at double its working voltage to insure this unfailing performance.

> THE ELECTRO MOTIVE MFG. CO., Inc. Connecticut Willimantic,

MANUFACTURERS ARE INVITED TO SEND FOR SAMPLES



FOREIGN RADIO AND ELECTRONIC MANUFACTURERS COMMUNICATE DIRECT WITH OUR EXPORT DEPT. AT WILLIMANTIC, CONN. FOR INFORMATION. ARCO ELECTRONICS, INC. 103 Lafayette St., New York, N. Y.—Sole Agent for Jobbers and Distributors in U. S. and Canada



How to discipline fuel—in flight

It takes more than an experienced pilot and well engineered control surfaces to keep an airplane in straight and level flight.

Fuel must be delivered to the engine without fail. The fuel must be controlled—prevented from sloshing from one cell to another or from end to end of the cells.

That's where the little flapper valve, manufactured for the B. F. Goodrich Company, comes in Placed in fuel cell baffles, and opening inboard, these devices keep the fuel in place despite any changes in flight attitude, permit the fuel to flow only in the proper direction.

They must be positive in operation. The material from which they are made must be inert in the presence of "loaded" aviation fuels, dimensionally stable, non-absorptive, light in weight . . . and, for reasons of economy, be easy to manufacture.

All these requirements add up to a call for Synthane molded-macerated plastics. But there is much more to the story. Synthane has a wide and rare *combination* of chemical, mechanical and electrical characteristics which make it a material for almost all industry.

It is hard, dense, exceptionally strong for its weight. Synthane is an excellent electrical insulator, has low dielectric constant and low power factor. It is economically produced in molded forms of moderately complex shapes. Parts may be quickly machined from Synthane sheets, rods or tubes on standard equipment. It is moisture and corrosion resistant, and is thermo-setting.

These few of Synthane's many advantages may suggest its value to you. If you would know more about Synthane, send for the Synthane Catalog. Synthane Corporation, 6 River Road, Oaks, Pennsylvania.

PLASTICS WHERE PLASTICS BELONG



When you need a FUSE-think of BUSS

THE MOST COMPLETE LINE OF FUSES FOR THE

Electronic Industries

- RADIO
 - TELEVISION
 - RADAR
 - INSTRUMENTS
 - CONTROLS
 - AVIONICS

Whatever your needs in electrical protection there's a Buss fuse made to fit.

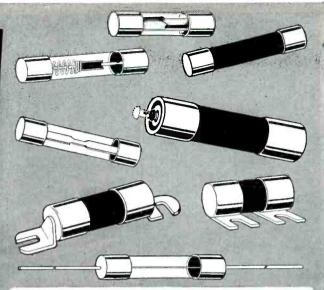
Send for Bulletin SFB—showing full line of fuses and fuse mountings.

SPECIAL FUSES, FUSE CLIPS, FUSE BLOCKS and FUSE HOLDERS

Sometimes a special fuse or fuse mounting is required. In such cases we welcome your requests either to quote—or to help in designing or selecting the special type of fuse or fuse mounting best suited to your particular conditions.

Submit description or sketch, showing type of fuse to be used, number of circuits, type of terminal, etc. If your protection problem is still in the engineering state, tell us current, voltage, load characteristics, etc.

At any time our staff of fuse engineers is at your service to help solve your problems in electrical protection.



A complete line of fuses made to dimensions smaller than National Electrical Code fuses.

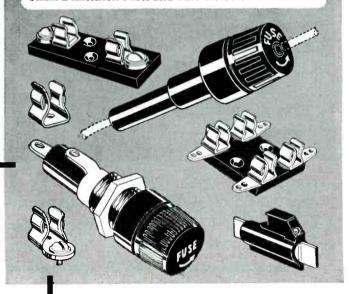
These fuses are SPACE SAVERS.

They are particularly well suited to the protection of instruments, radios, television and electronic equipment of all kinds, aircraft, automobiles, coin-operated devices and any apparatus where space for the protective device is at a premium.

Fuses of the Dual-Element, Renewable and One-Time type are available.

Companion lines for FUSETRON and BUSS small dimension fuses are BUSS Fuse Clips, Blocks and Fuse Holders. They are made in many types and sizes to make it easy to select the fuse and fuse-mounting needed to give the required protection.

For full information ask for the BUSS Bulletin on Small Dimension Fuses and Fuse Holders — Form SFB.



USE THIS COUPON - Get all the facts

Bussmann Mfg. Co., University at Jefferson St. Louis 7, Mo. (Division McGraw Electric Co.)

Please send me Bulletin SFB containing complete facts on Buss Small Dimension Fuses.

Title Company

Address

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City_____State____



"In" at the start ... Since that "eye-popping" day of aviation's first successful blind flight, Kollsman has specialized in the making of precise, dependable instruments. > In the evolution of modern flight, Kollsman was literally "in" at the start. For a Kollsman Sensitive Altimeter made accurate instrument flying more than a laboratory "pipe-dream." * And later - to pace the needs of aviation science -Kollsman developed a series of complex miniature motors — units to expertly control involved systems of instrumentation. ** Today, these high-precision, steady-working Kollsman motors give new meaning to accuracy and control in countless numbers of applications.

for precision and dependability look to KOLLSMAN

Elmhurst, New York



... about

custom designed trimmers

Pictured above are several custom designed trimmers that incorporate the elements of standard Erie Disc and Tubular Ceramicon Trimmers. Each has been developed for a specific purpose, and each does its job efficiently and economically. Proper design and precision manufacturing, plus our years of experience, are the keynote to Erie quality.

Look at these units carefully. They should suggest the possibility of using Erie Resistor know-how and facilities to make your equipment more compact and more efficient.

Erie has the most complete trimmer line in the industry. We would like to work with you on combining trimmers, fixed capacitors, and other circuit elements into integrated sub-assemblies. Inquiries should specify complete mechanical and electrical requirements.

- Special ribbon type terminals on standard Style TS2B Trimmer for direct connection to other components.
- 3 Compact Trimmer—Capacitor—Resistor
 —Coil Design. A complete oscillotor
 unit.
- Where special mounting is desired, standard Erie Style TS2A and Style 557 Trimmers con be supplied mounted on brackets.
- 7 Two trimmer elements become an integral part of this coil form and I. F. top section,
- Special bracket and terminal arrongements on dual trimmer unit.
- A compact pluggable ossembly for mounting a trimmer in parallel with a plug-in crystal.
- (11) Special tubular ceramic trimmer and variable inductance having one common terminal.
- (12) Special steatite tubular dual trimmer.
- (13) Standard Erie Style 557 Trimmer with special bent rotor terminal.

Electronics Division

ERIE RESISTOR CORP., ERIE, PA.

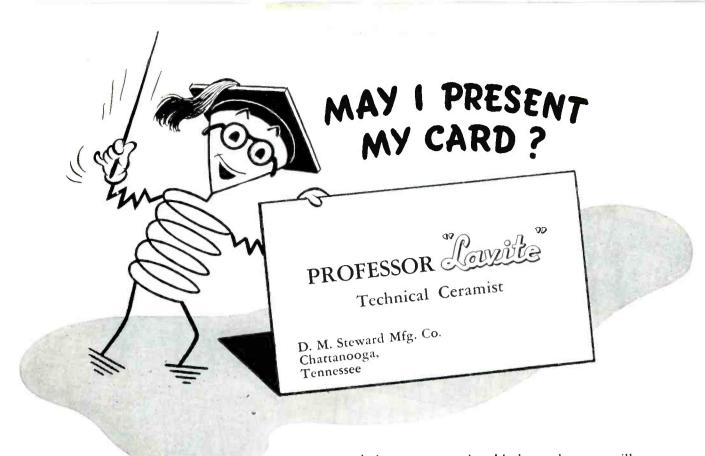
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Beden WIREMAKER FOR INDUSTRY





Ask for your copy of our booklet, "Lavite" Technical Ceramics, Table of Properties. (Steatites, Titanates, Ferrites and others.) take me into your confidence on all those knotty problems involving technical ceramics. You see, "Lavite" is more than just my name—it is the name by which the wide selection of Steward Technical Ceramics ("Lavite" Steatites, "Lavite" Titanates, "Lavite" Ferrites and many others) have become widely known. In fact, I am truly flattered to know that "Lavite" has become the name and standard that many of the most exacting users of technical ceramics insist on and specify religiously. "Lavite" Ceramics are not new developments—in fact, they have been in research and redevelopment for 75 years. "Lavite" Ceramics offer you practically any desired combination, or balance, of physical properties to satisfy your needs—from ordinary radio and general medium-frequency applications to those demanding precise qualities of low electrical loss, high mechanical strength and thermal shock resistance, plus the advantages of excellent machineability which permits production of your ceramic parts to closest tolerances.

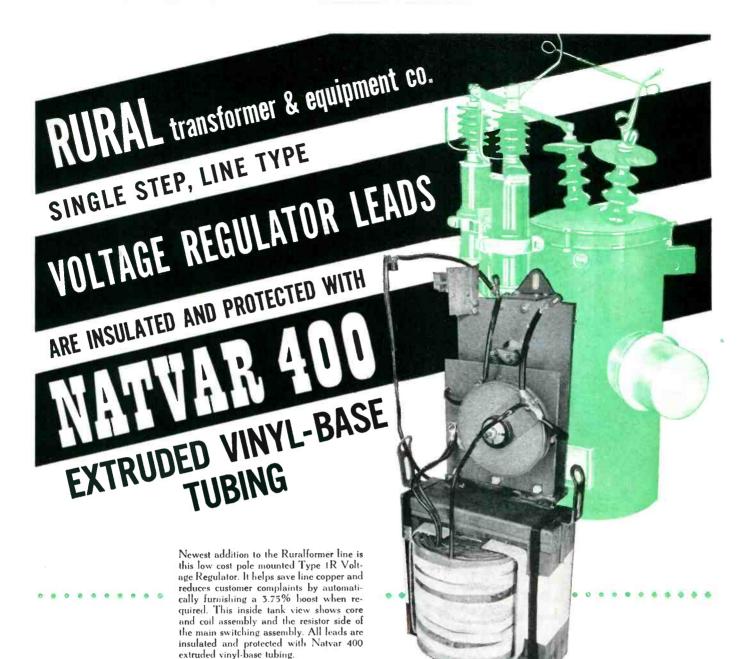
Send me your specifications for practical and proved recommendations, or ask for a testing sample of any one of this famous family of technical ceramics without obligation.

D. M. STEWARD MANUFACTURING CO.

3604 Jerrome Avenue Chattanooga, Tennessee

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The Rural Transformer & Equipment Co., Milwaukee, manufactures distribution transformers and associated equipment. They have developed a low cost, high surge strength transformer for distribution of rural power at 14,000 volts. With externally operated tap changer, bonded coils, and externally mounted protection against lightning surges and overload, Ruralformers carry a strong service guarantee.

Component parts and materials have been carefully picked. Natvar 400 extruded vinyl-base tubing is used to insulate and protect leads because of its consistently superior resistance to both oil and heat.

It will pay you to use Natvar 400 tubing, tape, and other Natvar flexible insulating materials for your requirements. They are available either from your wholesaler's stock or direct from our own.

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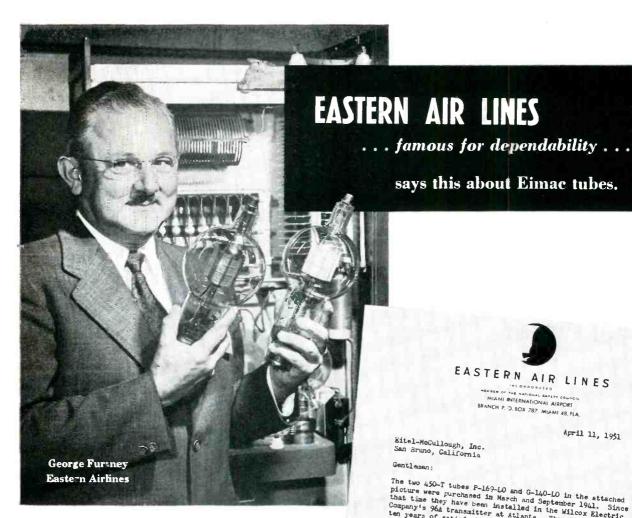
NATVAR: Rohway, N. J.

201 RANDOLPH AVENUE * WOODBRIDGE, NEW JERSEY

Natvar Products

- Varnished cambric-straight cut and bias
- Varnished cable tape
- Varnished canvas
- Varnished duck
- Varnished silk
- Varnished special rayon
- Varnished Fiberglas cloth Silicone coated Fiberalas
- Varnished papers
- Slot insulation
- Varnished tubings and sleevings
- Varnished identification markers
- Lacquered tubings and sleevings
- Extruded vinyl tubing and tape
- Extruded vinyl identification markers

Ask for Catalog No. 21



The statement by George Furtney of Eastern Air Lines once more proves that the workhorses of modern commercial communication systems are the Eimac 450T type triodes. Their dependability is unchallenged.

The 450T is rated at 450 watts plate dissipation, 6000 max. plate volts, and 600 ma. max. plate current at frequencies as high as 40 Mc. They are widely used as either amplifiers, oscillators, or modulators. Complete data and application notes will be furnished upon

The 450T type triodes and other outstanding Eimac vacuum tubes are today specified in ever increasing numbers to fill key sockets in important military, communications, and industrial equipments. Consider them for your equipment . . . they have been proved in service . . . an unbiased proof that is your assurance of undeniable dependability.

EASTERN AIR LINES

says this about Eimac tubes.

HENRE OF THE NATIONAL BAFFIY COUNCIL
MIAMI BYTERNATIONAL AIRPORT
BRANCH P. D. BOX 787, MIAMI 48, FLA.

April 11, 1951

Eitel-McCullough, Inc. San Bruno, California

The two 450-T tubes P-169-LO and G-140-LO in the attached picture were purchased in March and September 1941. Since they have been installed in the wilcox Electric Company's 964 transmitter at Atlanta. They have given almost use,

The record of these tubes, together with 450TL F-113-LO just removed from service proves that our choice of EDMAC tubes was correct for dependable operation in the Aviation Safety Services.

Very truly yours, Kafiertury G. W. Furtney Engineering Assistant to Director of Communications

GWF:fd

THERE'S NO SUBSTITUTE FOR EASTERN'S EXPERIENCE

A new Eimac quick-reference catalogue is now available. Write for yours.

Eitel-McCullough, Inc. Bruno, California Export Agents: Frazar & Hansen, 301 Clay St., San Francisco, California

290

Button, Button, who's got the Button (Capacitors)?



Sangamo can furnish SILVERED MICA BUTTON CAPACITORS

for any requirement...

You can look to Sangamo whenever you need small sized, light weight button capacitors, with extremely low series inductance, for application in high frequency circuits. Sangamo Button Capacitors are encased in silver plated, corrosion resistant brass and are exceptionally stable over a wide temperature range. The case is utilized as the low potential terminal.

Sangamo Silvered Mica Button Capacitors meet all requirements of V.H.F. and U.H.F. applications. They can be furnished with temperature coefficients and drift characteristics up to and including "E" of Armed Services specification JAN-C-5. The normal operating temperature range is from minus 50° C. to plus 85° C. Higher operating temperature requirements will be considered and may be negotiated with The Sangamo Capacitor Engineering Department. Catalog No. 830 gives full details—write for your copy.

Your Assurance of



Dependable Performance

ELECTRIC COMPANY SANGAMO

SPRINGFIELD, ILLINOIS

IN CANADA: SANGAMO COMPANY LIMITED, LEASIDE, ONTARIO

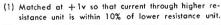
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FOR READY REFERENCE...

A Complete List of lelded GERMANIUM DIODES

LL of the General Electric diode products listed below have the following advantages over other types of rectifiers commonly used: welded contact...low shunt capacity...insulating case...small physical size...long life...no filaments... no contact potential...self-healing...high moisture resistance. Complete bulletins sent on request. Write: General Electric Co., Sec. 471, Electronics Park, Syracuse, N. Y.

CATEGORY	RTMA DESIG- NATION	G-E TYPE	PEAK INVERSE VOLTAGE	CONTIN. OPER. INV. VOLTAGE	MIN. FORWARD CURRENT (MA) AT +1V	MAX. INV. CURRENT (@a) AT 50V	AV RECTIFIED CURRENT (MA)	PEAK RECTIFIED CURRENT (MA)	SURGE CURRENT (MA)	SIZE
	1N48	G5	85	70	4.0	833	50	150	400	A
	1N51	G5C	50	40	2.5	1667	25	100	300	A
GENERAL	1N52	G5D	8.5	70	4.0	150	50	150	400	A
PURPOSE	1N63	G5E	125	100	4.0	50	50	150	400	A
	1N75	G5M	125	100	2.5	50	50	150	400	A
	1N64	G5F	20	20 Min. dc current in 44 mc rectifier—100@a						A
TV	1N65	G5G	85	70	2.5	200	50	150	400	A
	1N69	G5K	75	60	5.0	850	40	125	400	A
JAN	1N70	G5L	125	100	3.0	410	30	90	350	A
VHF		G6	15 Min. Rect. Eff. at 100 mc and 2 v signal—60%							A
	1N72	G7	(+3db max	noise factor o	ver 1N21B in) 25	75		В
	11.72	G7A	500 mc mixer CKT 25 75							A
		G7B	5 (5 () 25 75						В
	G7C 5 75% min, rec., eff. at 100 mc for detector 3 25					\$ 25	75		A	
UHF					5 (Tested for sharpness of break E-I char, for) 25					В
		G7E	5 7	freq. multiplier \$ 25 75						A
1	1	G7F	5 \	1000			25	75		В
		G7G	5 {	60% min. rec'. eff. at 100 mc for detector \$ 25 75						A
		G8	85	70	4.0	833	50	150	400	A
MATCHED		G8A	8.5	70	4,0	150	50	150	400	A
PAIRS		G8B	125	100	4.0	50	50	150	400	A
Note (1)		G8C	125	100	2.5	50	50	150	400	A
QUADS	1N73	G9	75		Note (2)	50@ —10v 50@ —10v		60	100	D D
aches.	1N74	G9A	75		Note (3)	30@ -100	22.5	- 00		
TRANSISTORS Note (4)		G11 Amax. RMS emitter signal level—.3v; max. DC emitter current—.5 ma; Max. DC collector current—2.0 ma; power gain 17 to 21 db; power output 25 mw					c c			



⁽²⁾ Consists of 4 balanced diodes. With 15 ma forward current; the voltage drop of each diode is 1.3v min. and 1.7v max., all diodes are within 0.1 volt of each other, and voltage drop of a pair is 0.03 volts of each

⁽⁴⁾ Additional test over G11 for negative resistance of base current vs. base voltage characteristic for trigger circuit operation.



You can put your confidence in

GENERAL



ELECTRIC

⁽³⁾ Consists of 4 balanced diodes. With 15 ma forward current; the voltage drop of each diode is 1.2v min. and 1.8v max., all diodes are within 0.2v of each other, and voltage drop of a pair is 0.1v of each other.



BUSINESS BRIEFS

By W. W. MacDONALD

Washington's Attitude during the last war appears to us to have been "take care of the assemblers of complete equipment and they will take care of the component makers." It has been quite a while developing, but this time we think we see signs that it will be "keep component makers in business during the unsettled period and let assemblers paddle at least part of their own canoe."

Some time will elapse before this new thinking is translated into action that all component makers can feel. Action seems, however, to be definitely on its way.

Another Trend we see developing in the nation's capital is a long-overdue shift from the "study and statistics" phase of the mobilization effort into the "action even if every possible contingency cannot be foreseen" stage. Personnel changes within several government agencies are partially responsible for the trend. In particular, association of more men from our industry with government agencies, on a part-time or full-time basis, is responsible.

Most Important Date for our industry is July 1, when the Controlled Materials Plan goes into effect.

CMP is designed to insure adequate supplies of critical materials for essential defense and defense-supporting projects. It leaves manufacturers of nonessential civilian items out in the cold to scramble for any materials that are left over. The pickings may be lean, or they may be lush.

Just what CMP will do to civilian equipment production cannot be evaluated in advance because the situation is quite different from that during World War II. Then, production facilities were heavily engaged on military equipment and no one worried much about civilian equipment production. Now, production facil-

ities are only partially devoted to military equipment and civilian equipment production is important in the economy, particularly the production of television sets.

Several government officials to whom we talked in Washington just before press time thought there might be enough materials available in the open market for the balance of the year to support a healthy civilian business, particularly if the use of noncritical substitutes expanded. Others thought it might eventually be necessary to bring certain materials required for the production of civilian goods under CMP.

Munitions Board is currently inviting responsible electronic-equipment manufacturers and their technical personnel to look at gear recently designed and built for the Air Force and the Navy. Such gear is being exhibited, despite the fact that it is classified, in certain major cities with the hope that by examining it new government contracts will be more acceptably executed.

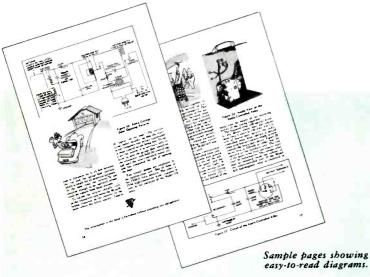
We looked over the equipment recently exhibited in New York, and while identification cannot be made without violating security regulations, we can tell you that some of the construction, components and materials opened our eyes. If you have an opportunity to look over the equipment, by all means do so.

Elmer's Law is the one which says that if you are measuring a d-c voltage the initial arrangement of your voltmeter leads will be wrongly polarized.

Scientific Instrument Makers are unhappy about a recent tariff cut on certain imported apparatus. They point out that our government had to scramble for such apparatus when the supply was cut off from abroad during the last war and that an American industry virtually had to be started

Put Electronics to work

IN YOUR HOME!



HERE'S HOW TO MAKE 24 VALUABLE TIMEAND LABOR-SAVERS

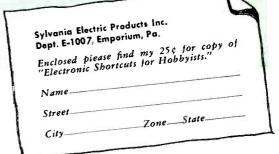
You don't have to be an electronics engineer to build these useful household gadgets. The step-by-step instructions in Sylvania's fascinating new book, "Electronic Shortcuts for Hobbyists," are written expressly for the home hobbyist, model maker and electrical experimenter. With this book you can build:

- A Radio-Controlled Door Opener.
- An Electronic Door Lock.
- A Charger for Small Dry Batteries.
- Radio-Controlled Relays.
- Pocket-Sized Stroboscopes.
- Remote Control for Model Trains.
- A Doorbell Booster.
- Photoelectric Relays.
- Photographic Interval Timers.
- An efficient Crystal Radio . . . and many other valuable gadgets.

All you need is some inexpensive Sylvania Crystal Diodes and a few everyday materials. Book contains full instructions and easy-to-follow diagrams. Send a quarter along with the coupon for your copy.









SYLVANIA ELECTRIC

ELECTRONIC DEVICES; RADIO TUBES; TELEVISION PICTURE TUBES; ELECTRONIC TEST EQUIPMENT; FLUORESCENT TUBES, FIXTURES, SIGN TUBING, WIRING DEVICES; LIGHT BULBS; PHOTOLAMPS; TELEVISION SETS

SHOCK and VIBRATION NEWS

BARRYMOUNTS FOR ASSURED CONTROL OF SHOCK AND VIBRATION

SMALL AIR-DAMPED BARRYMOUNTS

for Miniaturized
Airborne Equipment

New-series Barrymounts, designed to meet requirements for compact isolators usable with miniaturized equipment, provide effective shock and vibration isolation in small space.



These mountings utilize air damping to minimize shock of aircraft landing and taxiing and to limit excursion so there is no snubber contact, even at resonance.



Upright and inverted types are available for two-hole or four-hole mounting. Unit mountings are one inch in diameter and 1-1/32 inches high under maximum rated load. Load ratings are 0.1 to 3.0 pounds per mount. The mountings weigh only 5/16 ounce each.





Bases using the inverted mountings raise the mounted equipment only 1/2 inch. Either upright or inverted unit mountings can be furnished on bases that conform to your specifications, load-ratings, and dimensions.

FREE CATALOGS

- 502 Air-damped Barrymounts for aircraft service; also mounting bases and instrument mountings.
- 509 ALL-METL Barrymounts and mounting bases for unusual airborne applications.
- 605-606 Miniaturized airdamped Barrymounts for use with airborne equipment.

STANDARD MOUNTINGS ISOLATE VIBRATION

Available for Aircraft, Marine, Mobile, Instrument, and Industrial uses.



Standard bases built to meet government specifications can be furnished by Barry; special bases can be supplied in sizes and load ratings to fit customers' exact requirements, including miniaturized bases. See catalog 502 and data sheets 605 and 606.





Aircraft vibration isolators designed to meet Army, Navy, and CAA requirements are available in ¼-pound to 45-pound unit ratings; also miniature mounts to 0.1 lb. See catalogs 502 and 509 and bulletins 605-6.

Instrument mountings are furnished for electronic components, tiny, fractional-HP motors, record changers, dictating machines, and other lightweight apparatus. See catalogs 502 and 504.





Shock mountings for mobile, railroad, and shipboard service also give vibration isolation at frequencies above 2000 c.p.m.; useful for general sound isolation. See catalog 504.

Industrial mountings isolate vibration from fans, motor-generator sets, transformers, punch presses, and other heavy industrial equipment. Bulletin 607 tells how to cut maintenance costs with Barrymounts.

from scratch. Now, they say, there is some danger that that industry may dwindle away. The other side of the picture is the desirability of supporting friendly European countries by supplying a market for their goods.

Here's another case in which it appears necessary for our government, faced by a half-war and half-peace situation that is unique in its history, to at least temporarily try to carry water on both shoulders.

Direct Quote much to the point is this one from a talk by GE's W. R. G. Baker before the Foreman's Association at Electronics Park:

"The real cost of mobilization must be paid by workers and . . . concerns engaged in production of civilian goods and services. The fewer persons and concerns producing civilian goods and services the less money there will be to finance military production."

Four War Years 1942 through 1945 saw our industry delivering 1,500,000 radio sets and 20,000 radar sets to the military.

Production of radio sets in the first quarter of this year totalled 4,235,597 and there were 2,199,669 television receivers made, according to RTMA. This represents increases of 27 and 37 percent over the first quarter of 1950. It is pointed out, however, that preliminary second quarter reports indicate a decline as against last year.

Some 95 percent of the tv sets made in the first quarter used picture tubes 16 inches or larger in diameter.

TV Receiver Inventory as of May 1 is estimated as between 1,750,000 and 1,850,000 sets by Television-Electronics Fund, Inc. of Chicago. This, according to the source, represents about three to three and a half months supply based on average monthly sales for the preceding 12 months.

TV Chassis intended for custom building into special cabinets or home recesses were bought by 90,-

THE BARRY CORP.

707 PLEASANT ST., WATERTOWN 72, MASSACHUSETTS

SALES REPRESENTATIVES IN

New York Rochester Philadelphia Washington Cleveland Dayton Detroit

Chicago Minerapalis *** Company
000 people in 1950, according to an estimate by John H. Cashman of Radio Craftsmen. Retail value of chassis is estimated at the \$10,000,000.

Government Purchases of radio transmitting and communications equipment during 1950 totalled \$116,477,892, according to RTMA. \$92,297,971 of the total was for radar. Purchases during 1949 totalled \$141,288,869.

North American Aviation says the present goal in automatic airplane guidance is a flight control system capable of directing an interceptor airplane through an entire combat mission. Takeoff, climb-out, target pursuit and destruction, and return to base would be accomplished automatically, with the human pilot merely selecting the target and monitoring the mission.

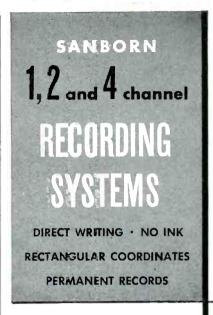
Once Before (p 60, Aug. 1950) we commented on the interest of horse players in electronic equipment. One of them had, we noted, used radio to send race results out of a park before said results reached the horse parlors, and so bet on sure things until dissuaded by the gendarmes.

Now the bookies are harnessing our art. One of them was picked up just the other day for interconnecting three successive telephones located at different places around New York City through automatic relays, making it difficult for the police to wiretap or raid.

Philco has 2.000 electronics field engineers serving with the armed forces all over the world.

This Month's Masterpiece of understatement: Electronics industry official, showing great restraint when discussing a governmentproposed change in broadcasting equipment operating procedure, said . . .

"We agreed to try the suggestion in conversations at Washington, and at Washington there wasn't much trouble involved. However, the nearer we approached the site of the work the more interesting the problem became."



RECORDERS AND AMPLIFIERS AVAILABLE SEPARATELY



In all Sanborn recorders, tracings are produced by a heated writing stylus in contact with heat sensitive, plastic-coated paper. The paper is pulled over a sharp edge in the paper drive mechanism, and the stylus wipes along this edge as it swings; thus producing records in true rectangular coordinates. The writing arm is driven by a D'Arsonval moving coil galvanometer with extremely high torque movement (200,000 dyne cm/cm deflection). Standard paper speed for the Model 51-600 recorder assembly, shown above, is 25 mm/sec. Slower speeds are available. Paper width 6 cm with 5 cm recording area. The assembly shown above is used in Models 128 and 141 (described above right) and provides the basic principles and methods on which recorders above right) and provides the basic principles and methods on which recorders for the 2- and 4-channel systems are designed



For complete descriptions, illustrations, tables of constants, and prices, write for catalog.

CHANNEL RECORDING **SYSTEMS**

Model 128 com prises a DC . General Purpose Amplifier in combination with the



Recorder Assembly shown below left, to which is added panel, transformer, and controls. Both instruments are contained in a single hardwood carrying case. The complete system is a vacuum tube recording voltmeter capable of reproducing in rectangular coordinates any electrical phenomena from the order of a few millivolts to more than 200 volts. When a Strain Gage Amplifier is specified, the system becomes Model 141. Amplifiers are readily interchangeable. When a built-in timer is included for either, the Model numbers become 128T or 141T.

CHANNEL RECORDING **SYSTEMS**

The two channels of Model 60 operate independently of each other, but record simulta-neously. Ten paper speeds are



standard equipment, in pairs of: 5 and 0.5, 10 and 1, 25 and 2.5, 50 and 5, 100 and 10 mm/sec. Ready interchangeability of amplifiers (DC and Strain Gage) and preamplifiers (DC and AC) makes possible the availability of a variety of input circuits. Timing and coding are built-in features. Each channel has a 5 cm recording width.

CHANNEL RECORDING **SYSTEMS**

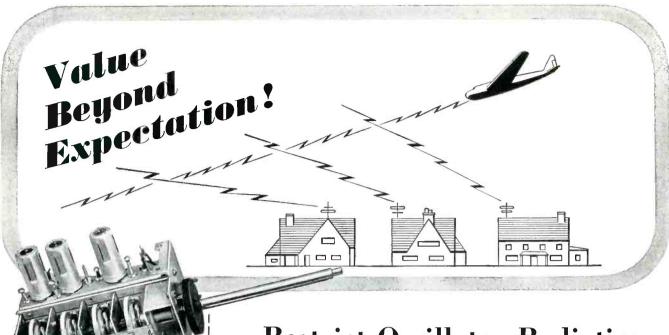
Model 67 provides for the direct, simultaneous registration of up to four phenomena on one record, using the same principles and methods as the two systems described above. In addition, there is a selection of eight paper speeds: 50, 25, 10, 5, 2.5, 1.0, 0.5, and 0.25 mm/sec., and provision for the use of 4-, 2-, or 1-channel recording Permapaper. As in Model 60, above, amplifiers and readily interchangeable.

preamplifiers are

INDUSTRIAL DIVISION

Sanborn Company

CAMBRIDGE 39, MASSACHUSETTS



Outstanding Advantages of the new Mallory Spiral Inductuner**

- A single control for easy selection and fine tuning of any television or FM channel.
- 2. Easily adapted to UHF converter use.
- 3. Excellent stability eliminates frequency drift.
- 4. Supplied in two-, three-, or four-section designs.
- Far more quiet operation; permits lower noise figures in front end designs.
- 6. Free from microphonics.
- 7. Greater selectivity on high frequency channels.
- 8. Eliminates "bunching" of high band channels.
- 9. Simplifies front end design and production.
- 10. Reduces assembly costs.
- 11. Choice of: 6 turn unit continuously tunes from channels 2 to 13; 4 turn unit tunes only channels 2 to 6, FM, 7 to 13; 3 turn unit tunes only the 12 television channels.

*Trademark:

***Reg. trademark of P. R. Mallory & Co., Inc. for inductance tuning devices covered by Mallory-Ware patents.

Restrict Oscillator Radiation with the Mallory TV Teletuner*

The development of a TV front end assembly which avoids interference with other electronic equipment is a recent and far-reaching accomplishment of Mallory research and engineering.

Built around the four-tuned-circuit Spiral Inductuner, this new front end is designed and constructed to limit radiation from the oscillator. Further precaution is taken by shielding the oscillator and converter from the RF amplifier. And, each section of the Inductuner is individually shielded. Thus, Mallory offers TV manufacturers a complete front end capable of performing within strict standards for oscillator radiation.

That's value beyond expectation!

The Mallory Teletuner is universally adaptable. It features higher gain and lower noise figure. It is available in 3, 4 and 6 revolution designs, incorporating the characteristics of the Spiral Inductuner listed at the left. Complete technical data sheets are available on request.

Television Tuners, Special Switches, Controls and Resistors

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Electrochemical Products
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Metallurgical Products

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July, 1951 — ELECTRONICS



CROSS TALK

▶ PICKUP . . . Last year about this time we irked our American friends in tv broadcasting somewhat by reporting that British tv stations do a conspicuously better job of televising film than we do, despite the fact that the British video band is 1.25 mc narrower than the American. Some of those who doubted the report have since been to London to see for themselves and have confirmed the superior quality of British film transmissions. The explanation, one of the reasons advanced in favor of the European scanning standard of 25 frames per second, is the fact that flying-spot scanning can be used, with this standard, merely by running the film about 4 percent faster than normal. With the American standard of 30 frames per second this is not feasible, so we use storage-type cameras, notably the iconoscope, to perform the transfer from 24 film frames to 30 scanning frames per second.

Sporadically reports have appeared that a new scanner was in process of development that would overcome the impasse and permit flying-spot scanning of film on the American standard. Such a film scanner was, in fact, shown to the SMPTE a few weeks ago by the Bell Telephone Labs (p 114, this issue). The new device uses two clever electronic servos to overcome vertical jitter and flicker, and pro-

duces a picture that is very, very good—better than anything on the air today anywhere. For various reasons, Ma Bell does not intend to produce the scanner commercially, but is willing to license others to do so. Our guess is that, when the development is complete, several large corporations will be competing for the privilege.

► SECOND CHANCE . . . In view of the great importance of education in our field, we note with pleasure the establishment of the Majestic Research Fund, Inc. which will provide 48 "second-chance" scholarships per year, valued at about \$2,000 each, to persons over 35 years of age who were unable to secure or complete their formal education in earlier years. fund of \$500,000 was set up by Majestic Radio and will be administered by a board of officials of NYU, Columbia, Brooklyn Polytech, Textile Workers ofthe A worthy project, and America. one which can be copied by other companies without fear of committing infringement.

The growing realization of industry's stake in higher education was voiced recently by Alfred P. Sloan, Chairman of General Motors, at a dinner signalizing his gift of \$5 million to found a school of industrial management at MIT. Mr. Sloan pointed out that it is becom-

ing impossible to operate colleges and universities by the investment of endowment funds, and that no relief from this difficulty is in sight. He suggests that industry at large ought to set up a foundation to make outright grants to such institutions, for education and research, without expecting direct monetary return, the funds to be liquidated over periods as short as five years. Continuing support of a given research project would then depend on actual results and potential value. The big foundations and many individual companies are already doing this, but the base has to be very broad to support a program as big as this country needs.

▶ ODE . . . As we write, the Supreme Court has just rendered its opinion on the color television case of RCA vs FCC. Also, within a few days we take our departure for Geneva, for the Sixth Plenary Session of the C.C.I.R. So, all in all, we think it safe to print the following ditty, which is not original with us or Rudyard Kipling:

The filters and dichroics die,

The mirrors and the disks depart

Still stands our basic raster type And readings on the testing chart.

Dear FCC, be with us yet,

May we forget, may we forget!

Production Planning

The six post-war years have seen per-man-hour production of electronic equipment more than double, at the Sperry Gyroscope Company, despite increased complexity of products and an increase in the unskilled-to-skilled operator ratio. Details of production planning program

that made this increase possible are presented

PRODUCTION of electronic equip- in fact, the motivating influence ment reached an all-time high during World War II. Some of the methods used to attain these high production figures were of necessity inefficient, but they were nevertheless effective. One fortunate aspect of the present state of affairs is the fact that industry today has an opportunity to plan ahead. While there is no certainty of another global conflict, there is a threat, and by laying careful plans now, the occurrence of such a disaster can be met with methods that are efficient, as well as effec-

At the Sperry Gyroscope Company, production per man hour of military electronic equipment has more than doubled since the end of the war. This is on an average basis-in some instances increases of 8 and 10 have been realized. This statement is even more impressive than it appears at first because of the increased complexity of the equipment being turned out today. Also adding to the impressiveness is the fact that production progress is being accomplished with the use of less-skilled help. The end products, again as a result of planning and specialization, are of better quality. Quality improvement was,

that started the program that made the increases possible.

The Sperry System

Many of the new ideas that enable a company of Sperry's size to increase production come about through paper work and the shuffling of personnel both on the management side and in the shop.

Key to the system is the lettersized operations sheet. These sheets break each job down into step-bystep functions. Descriptions of these various functions are in plain language, written so that operators even in the lowest skill categories can understand them and carry out the prescribed operations.

The use of such sheets offers many advantages:

- (1) Because of the great detail used on operations sheets, in contrast to that which could be placed on a blueprint without undue complication, greater quality and uniformity of products is obtainable.
- (2) An operator who has used one sheet can follow similar sheets almost without further instruction, which relieves shop foremen for other duties. This also enables the production department to run one job for a time and then to switch



FIG. 1-Setup for cutting uniform wire lengths. Shear jaws are actuated when wire end comes in contact with switch



FIG. 2—Old method of removing shields involved snipping with shears. New methods are shown in Fig. 3 and 4

to another without lost motion.

- (3) Fewer mistakes are likely to occur when an operator has only to follow a description of his particular job, rather than an over-all blueprint. In fact, little knowledge of blueprints is required—instructions are in words and sketches.
- (4) The sheets, being of small and uniform size, are readily filed and easily available for referencethe unfolding of monstrous blueprints is always time consuming and distracting to others.
 - (5) If production on a particular
- DURING WORLD WAR II, approximately 40 man hours were required at Sperry for turning out a particular amplifier chassis, and the average skill classification was B.
- Today, as a result of production planning and technological advancement, the same amplifier chassis is completed in 6 man hours, and the average skill is D. (Skill D is one step above trainee!)

Speeds Military Orders



FIG. 3—Production shield-removing operation. At left, operator forms pyramid from shield. Co-worker removes pyramid with nibbling machine so that shield end falls free

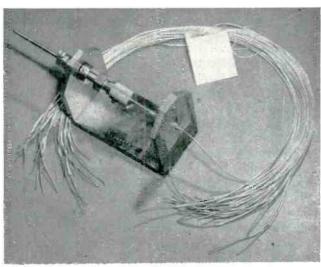


FIG. 4—New shield remover involves one simple operation.
Use of this tool is explained in text. Note wire tag containing instructions attached to bundle



FIG. 5—During World War II, harness making became an important adjunct to equipment fabrication. This photograph shows a typical installation from that period

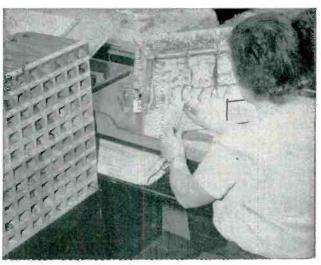


FIG. 6—Since the war, Sperry has developed and adopted the wire-rack shown at extreme left for holding wires in proper sequence of use in harness

job must be stepped up, the operations sheets may be further subdivided so that each worker does only half of the steps, thereby speeding the operation without additional training and indoctrination. Extra workers are simply sandwiched in between the regulars and given their share of the operations.

Each job has its operations sheets. In fact, at Sperry, as many as 175 different varieties of products may be turned out in a single month. In the accompanying illustrations, it will be noticed that

most of the workers have their operations sheets conveniently located for quick reference—unless their jobs are of a nature where reference is unnecessary.

Production Procedure

The wire preparation department at Sperry provides a good example of recent improvements. Looking back, the day is not too far gone when a chassis was wired by a skilled technician armed with a coil of wire, a pair of needle-nosed pliers and a soldering iron. His

knowledge had to include complete familiarity with schematic diagrams, or drawings, and experience with soldering, identifying components and contact numbers, cabling and lacing, and it was desirable that he appreciate the importance of lead dress and circuit functions.

That type of operation more or less disappeared in the last war, but significant improvements have been made even in the past five

Today all wires are prepared in bundles prior to assembly. A small tag is the start of each bundle. On receipt of instructions from the production department, a shop worker refers to a list of wires that will be needed for a given job. A tag is then made out for each wire in the job. This tag tells the wire preparation department all it has to know to prepare the bundle. Length, type of insulation, gage of wire and end preparation details are included. Special instructions are given to shielded wires, telling what length of outer insulation is to be removed, where the shielding is to be cut, whether or not a ground wire for the shield is required, and so on.

The completed tags then follow each wire bundle through cutting, stripping, tinning and any other operations, so that the worker at each operation can abstract his instructions without reference to anything but the tag. Tags are visible in many of the accompanying photographs. Similar orders simultaneously go to all other production departments, with the result that necessary components converge on the assembly area at just the right time. Small parts are parcelled into convenient stack bins and numbered in sequence in the order of their use in the particular iob.

Finally, appropriate operations sheets are dispatched among employees. Each employee goes to the appropriate departmental storage area, withdraws his necessary equipment, and production begins without delay.

Wire Preparation Details

When a tag is issued for a batch of wires, it first goes to the cutting department. Small sizes are then cut to length and stripped on conventional Artos automatic wirepreparation machines. Special lots are allocated to either hand operations, or special machines of the type shown in Fig. 1. This whydidn't-I-think-of-that idea is typical of aids that can drastically cut down on production time and increase efficiency. It consists simply of a pair of pneumatically-operated cutting jaws, a yardstick, and a snap-action switch which slides along the yardstick. The heavy wire, which cannot be processed by

the Artos machines, is pulled from a coil through the cutting jaws until its end comes in physical contact with an improvised target that is fixed to the snap-action switch. At that instant, the jaws cut the wire to the desired length. A pointer on the yardstick permits adjustment for different lengths.

From the cutting operation, the smaller wires go to the twisting and tinning operation. Several other time-saving operations are applied at this step, where larger wires—especially shielded cable—are processed.

Figures 2, 3 and 4 show the old and two new ways for removing outer shielding. The old way, and this method is now in use at many plants, involved an operator using a pair of shears to remove the prescribed amount of shielding. first of the new systems (Fig. 3) consists of two processes, each handled by separate operators. The first is referred to as pyramiding. The operator grasps the shield with two pairs of V-shaped jaws. He then bunches the shielding by pushing the jaws toward each other, with the result that a pyramidshaped flange of shielding forms around the wire. The wire, thus prepared, is then passed to the next operation, which is known as nibbling. Here a pair of flush cutting surfaces nibble away at the raised portion of the shield as the operator turns the wire, Then the shield between the end of the wire and the removed pyramid drops free.

The second shield remover is a recent addition, and is still undergoing evaluation. The operation is much simpler and faster. The tool is shown in Fig. 4. A hardened steel tube, attached to a universal joint, passes through a circular inside cutting surface. The tube's inside diameter is sufficient to accept the insulated inner wire. while the shielding fits over the tube. It is relatively simple to push the wire onto the tube in this fashion. After inserting the wire, the tube is moved circularly around the inside cutting surface of the tool. The blade shears the shield in several motions and leaves the shield conveniently flared for insertion of a ferrule. The steel tube is held in place by a chuck to facilitate rapid changes for different wire sizes. A graduated rod in the tube measures the amount of shielding to be removed.

Most marking is done as a matter of manufacturing convenience, though some jobs call for marked leads for certain terminal board connections where marking facilitates service in the field. Commercial markers may be used, or small pieces of premarked tubing fitted over the ends of the wires. In the latter case, a slight bend in the end of the wire will hold on a piece of rigid tubing until the lead is soldered in and the marker removed.

It was found in using commercial marking machines that a conventional typewriter ribbon would serve well as a source of ink, if the imprints were subjected to a drying process. An ordinary 200-watt infrared lamp shining on a boxfull of pieces of tubing does the job of drying thoroughly and quickly.

Conventional automatic and manually operated lug and ferrule machines insure good connections to wires at their terminations.

Harness-Making Hints

The value of prefabricating wire harnesses for complicated chassis has long been known. Harness boards save countless hours in production time. By prewiring multiconnection plugs before the harnessing operation, even more time is saved. Dugouts must be provided to accept the plugs in the proper plane. Figures 5 and 6 show typical old and new harnessmaking locations at Sperry. The job, shown in Fig. 6, involves a three-dimensional harness with plugs brought out at right angles to one of the planes. Cotter pins hold wire ends in place, while simple nails secure eyelets. Numbers at the ends of each wire enable the worker to proceed on a job, working from his operations sheet, with no experience on a particular model, but with certainty that what is done is right.

The wire rack at the left in Fig. 6 is a development that has generously rewarded its inventors. The wires are placed in the rack in advance, and in the order in which they are to be added to the harness,



FIG. 7—All plugs are prewired at locations such as these. Note convenient position of operations sheet for quick reference



FIG. 8—Subassembly location shows block of stack bins, solderingiron holder, rubber flap over a-c outlet and tool drawers

starting at the upper left and working down as one reads a page of written text. Each wire is precut to exact lengths and properly processed for the particular job. The wire racks are built of basic building-block rows of ten pigeonholes that may be placed on top of one another and in back of one another to accommodate more wires or longer wires.

Figure 7 shows a typical plugwiring station. Universal fixtures, used throughout the plant, and completely adjustable, hold one or several plugs. Notice the operator has her operations sheet at her left and the wire rack conveniently located. A spring fixed over the jig holds wires out of the way after they have been soldered in place.

Subassembly Prefabrication

Figure 8 shows a typical subassembly location. The stack bins are, as might be expected, all prefilled with the specified number and types of components, all placed in order of use, corresponding to the operations sheet for that job. The tool in the hand of the operator on the left is a wiping tool which bends the wire around lugs on terminal boards. The cradle type fixture permits rapid changes in position of subassembly.

The soldering-iron stand shown in the background has several convenient features. The iron may either be inserted from the back, as shown, for work requiring two hands, or from the front for removal and application to fixed assemblies. A specially-treated pad in front of the iron makes it easy for the operator to keep his iron clean, with resulting improvement in soldered joints. The chest of drawers under the table is provided for the employees' tools and personal effects. The use of these relatively simple facilities greatly reduces cluttering of the working area with cosmetics, cigarette packages and so on. A clip-on rubber protector over the 110-volt outlet under the bench prevents down time as a result of loose wire ends shorting out power lines and eliminates certain human hazards.

The leads of all small parts are preformed, prior to assembly, on special machines designed for that purpose. When a resistor arrives at the assembly area, it has already been trimmed and bent to conform to the points between which it is to be soldered. Short bus pieces are also thus preformed.

Inspection locations are dispersed through the plant so that rejects take the form of small subassemblies instead of completed units.

The plant is at present in the process of installing swinging ash trays that may be moved out of sight under the bench when not in use. This reduces the possibility of ashes getting into work, and adds generally to the appearance and efficiency of the shop.

Soldering irons are assigned to

each individual operator. Simple hand tools are his personal property; these are made available at cost to him through the plant tool room. This scheme introduces an actual improvement in work through pride of personal ownership.

It was found that ten minutes of production time was wasted at the end of each shift because the operators pulled their plugs to let their soldering irons cool for storage. This waste was eliminated by providing asbestos scabbards for the irons, which allowed them to be disconnected at quitting time and stored while still hot. Soldering iron tips are maintained by a special crew. Special partly-chromium-plated tips are used.

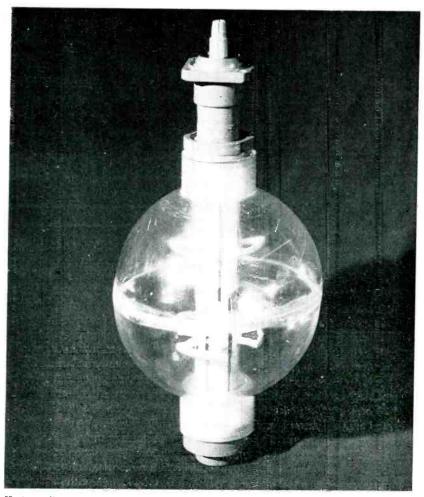
Future Planning

The basic operations outlined are, for the most part, typical of the things that make increased production efficiency possible. They are mostly simple things, brought about by cooperation and understanding and a will to turn out good work efficiently. Their incorporation was not a matter of chance, but the result of a five-year study of methods and facilities.

Carried through other assembly operations, the same kind of planning that has facilitated wiring is making it possible for less-skilled production workers, backed up by experienced foremen working with a highly skilled group of production engineers, to do a better job.—J.D.F.

Horizontally-Polarized

Provides azimuth ratio less than 3 db for frequency range of 2,970 to 3,125 mc with good voltage standing-wave ratio. Vertical pattern has half-power points at 45 degrees, which gives antenna a gain of approximately 2 db over half-wave dipole



Horizontally-polarized omnidirectional antenna for 10-cm region mounts directly on rigid %-inch coax. Plexi-glas ball permits pressurizing antenna

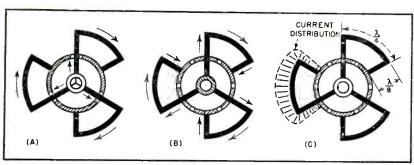


FIG. 1—Configurations shown at B and C are improvements of that shown at A. Addition of balun slots improves electrical and mechanical characteristics

By CHARLES BRASSE, JR. and RICHARD THOMAS

Bendix Radio Division Bendix Aviation Corp. Baltimore, Maryland

HE ANTENNA described herein was designed for use in the 10-centimeter band, and the particular application for which it was intended required that the azimuth ratio be less than 1.5 db at the center of the band.

To achieve a perfectly omnidirectional pattern in the horizontal plane, with the polarization in this plane, the ideal radiating element should consist of a ring of uniform current. In practice, this type of current distribution may be approximated by means of a number of dipoles, slots, or loops arranged in a circular fashion and equally phased. All three types were considered in connection with the manner in which the elements could be mounted on a 7-inch rigid coaxial line, for convenience of manufacture, and it was decided that best results would be obtained through the use of loops as radiators.

Constructional Details

A very rugged feed system for loops can be easily built on the \$\frac{7}{3}\$-inch line by inserting an even number of radial rods equally spaced about the periphery of the coax, alternate rods passing through to the center conductor. Loops are then formed by joining the ends of pairs of adjacent rods, as shown in Fig. 1A, where three loops are arranged on six rods.

This model was constructed by

Omnidirectional Antenna

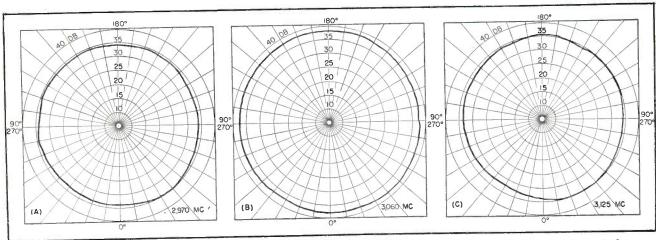


FIG. 3—Azimuth patterns show azimuth ratios of 1.25, 1 and 3 db for frequencies of 2,970, 3,060 and 3,125 respectively

using screws for rods, the ones joining the inner conductor passing through clearance holes in the outer conductor. This feed system can be improved mechanically and electrically by the use of slots as baluns and impedance-matching elements. Figure 2 shows a side view of this configuration, and Fig. 1B shows how the slots permit the proper polarities to be established. In this

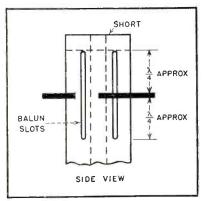


FIG. 2—Side view of configuration giving best results

model, the screws entering the inner conductor are also soldered to the outer conductor, thus forming a rigid support for the entire assembly.

The dimensions of the loops and the resulting current distribution are given in Fig. 1C from which it can be seen that the current is at a relatively high level at the rod ends. Dipoles could be used, but they would have to be longer, and the antenna would not be as mechanically rigid as the quarter-wave loops.

It would of course be possible to use the slots alone as radiating elements, by suitable probe feeding. In this case, however, the depth of the probes is somewhat critical and involves a tedious procedure of adjustment, especially if it is desired to stack two or more sets of elements to increase directivity in the vertical plane. By the use of loops, tuning is easily accomplished by increasing the length of the slots, which should be initially cut slightly less than a half wavelength.

A given set of loops will in general present an impedance which has low resistance and a capacitive reactance, and the negative shunt susceptance contributed by the slots will serve to match the antenna to the line with a minimum vswr of about 1.5. Increasing the loop perimeter will lower its apparent resistance and will produce a greater departure from a circular azimuth pattern, due to the shift of current distribution along the loops.

The photograph shows a twostack array which was enclosed in a hollow Plexiglas ball as a simple means of pressurizing the antenna. The upper loops are located 4 wavelength below the shorted end of the coax, and the lower loops, with feed reversed to obtain proper phasing,

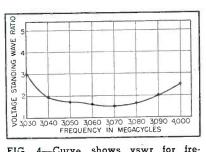


FIG. 4—Curve shows vswr for frequency range. No effort was made to extend range by broadbanding techniques

are spaced & wavelength below the upper set. The vertical pattern has a half-power beamwidth of 45 degrees, equivalent to a gain of approximately 2 db over a half-wave dipole. Figure 3 shows azimuth patterns taken at 2,970, 3,060 and 3,125 megacycles respectively, and the corresponding azimuth ratios are seen to be 1.25, 1 and 3 db. These patterns were taken on the Bendix automatic pattern-measuring equipment which provides a continuous recording covering a field strength range of 50 db, and incorporates facilities for self calibration in order to insure accuracy at all times.

The vswr characteristics of this antenna are shown in Fig. 4.

Only three loops were required for each element, and the resulting antenna combines satisfactory electrical performance together with mechanical simplicity, ruggedness and ease of adjustment.

Community CIVIL DEFENSE

Onondaga County, N. Y., which embraces the city of Syracuse, is setting up a system which utilizes the existing two-way radio facilities of 13 commercial and civic services, plus amateurs, and expects to have it in fully-coordinated operation by mid-August

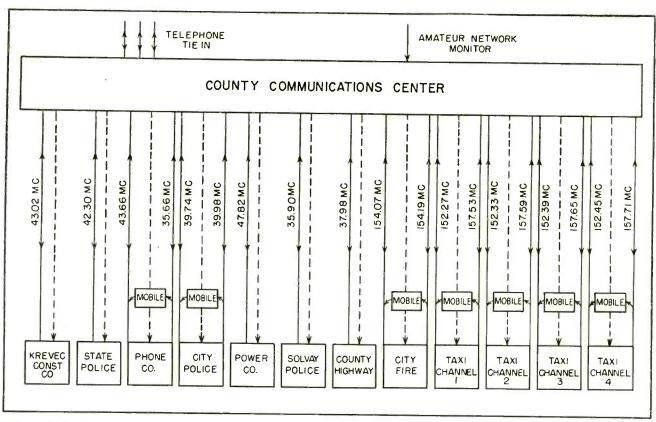


FIG. 2—Civil-defense system set up to make fullest use of available facilities and to provide an emergency backstop for them in the event of damage by sabotage or bombing. Dashed lines show paths of county-communications-center coordinating transmissions. Taxi-channel 1 is used by two independent companies

By NEAL F. HARMON

Civil Defense Coordinator Electronics Department General Electric Company Electronics Park Syracuse, N. Y.

Onondaga County, N. Y., including the city of Syracuse, are building a close-knit emergency communications system which will soon be ready for operation. The system will be flexible to the extent that it can supplement present communications in the area when needed, or take over for all present facilities in the event they are sabotaged or damaged in bombing attacks. It

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FIG. 1—Form used to determine twoway commercial radio facilities already available in Onondaga County

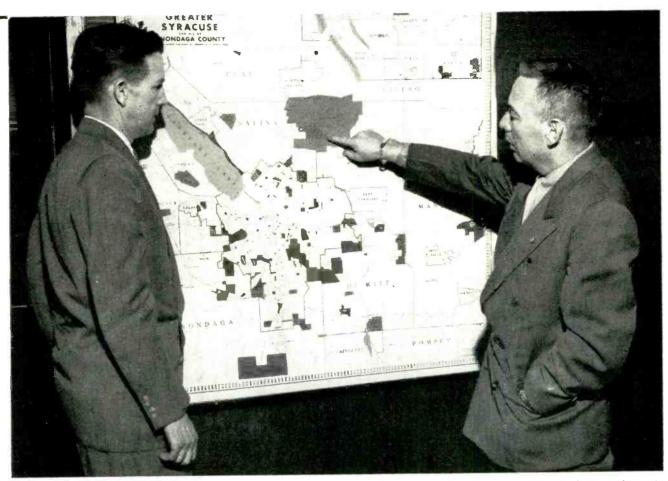
adds new radio facilities which can be used in peace as well as war emergencies.

Basically, the Onondaga County system revolves around a communications center equipped to (1) coordinate all existing commercial and civic two-way radio facilities in the area, (2) act as a standby system for any or all of these facilities if they are damaged in an emergency and (3) tie into other methods of communication such as the public telephone system and an amateur radio network.

Setting Up the System

Using the form in Fig. 1, a survey was made of local two-way

COMMUNICATIONS Plan



Author Harmon (left) looks over map of Greater Syracuse area preliminary to survey of civil-defense-communications requirements by local officials, with Harvey S. Smith, Onondaga County Civil Defense Director, at right



AMATEURS are fied into the system and their network monitored from the county communications center



POLICE provide an important part of the facilities

radio systems. Any community could utilize the same form to accumulate maximum information on available facilities. Using information recorded on the form a chart, Fig. 2, was made to pictorially present the entire system.

The survey revealed 13 services operating on 19 frequencies. These are the New York State Police, Syracuse Police, Village of Solvay Police, Syracuse Fire Department, Onondaga County Highway Department, telephone company mobile telephone subscribers, electric power company, a construction company, and five taxicab companies utilizing four frequencies. The County Sheriff's department shares the facilities and frequencies of the Syracuse Police Department. Seven of the services employ two-frequency operation,

with mobile units transmitting on a different frequency than the main-station transmitter. The 13 services operate more than 400 mobile units throughout the county.

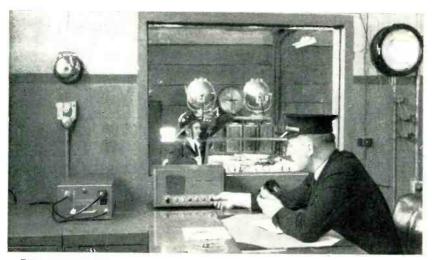
Communications Center

Civil defense authorities will maintain only supervisory control over the various services unless forced to take over for them because of loss of equipment, power or personnel. This means the CD center must be able to operate (1) as a mobile unit on any of the service frequencies, (2) as an auxiliary fixed station for any or all of the services should the need arise and (3) as a station from which to supervise and coordinate the entire system.

To do this, the center will be equipped with transmitters and re-

ceivers for each of the services. Seven transmitters will operate on two frequencies. There will also be a transmitter on a separate frequency for transmitting supervisory and coordinating messages, and a transmitter and receiver on the amateur frequencies.

Current plans call for the communications headquarters to be located in a large trailer (motor-truck-line type). Several remote locations will be established for tieing into telephone and power lines. As safety measures against possible sabotage, the center will be moved from time to time, and it will be unidentified insofar as call-letters are concerned. In addition to power connections at selected points, the trailer will be equipped with a gasoline-driven 110-volt a-c 10,000-watt generator.



FIRE DEPARTMENT radio equipment is incorporated into the overall network



TAXICAB COMPANIES operating mobile two-way radiophones cooperate with C-D



UTILITIES are tied in

It will be possible to operate all of the center's transmitters, and monitor transmissions for all of the coordinated services, from any one of four operating desks. Interlocks will prevent operation interference between desks.

When traffic is light, one operator will be able to perform all functions of monitoring, transmitting, and coordination. As traffic increases, a portion of it will be taken over by one or more additional operators at the other control desks. This division of duties will be conducted according to a predetermined plan.

Each operating desk will be equipped with key switches and relays to permit operation of all facilities. Each transmitter will be



TELEPHONE COMPANY mobile radio installations are incorporated in the plan



HIGHWAY MAINTENANCE radio equipment maintained by the county constitute an important link



CONSTRUCTION COMPANY maintaining its own private mobile radio system is part of the Onondaga County emergency network

protected by an interlock so that only one operator can operate it at one time. This plan affords maximum flexibility of control with a minimum of equipment.

"Alert" Receivers

Transmissions from the center on the coordinating frequency will be received at the dispatching points of the 13 services by means of an inexpensive receiver employing selective dispatch. All such receivers will be activated from the center by means of a tone-signalling system. They will, there-

fore, remain silent except to receive alerts and instructions from the center.

A tone-operated switch can be incorporated in these receivers, and used to activate warning and all-clear sirens directly from the trailer center. This will be extremely advantageous in some instances because the center will instantly receive alerts broadcast by the state-wide attack warning radio system.

All of the county's existing communication services are being encouraged to provide standby power equipment at their main stations. In addition, those services employing two-frequency operation are being asked to provide at least one of their mobile units with proper equipment to permit it to serve as an auxiliary dispatch station. Services operating on a single frequency need not alter a mobile unit to accomplish this but should have one mobile unit designated as an auxiliary dispatch station.

Four-Way Emergency Plan

Under the plan, there will be four possible dispatching conditions for each service. The first will be normal operation on regular power, then normal operation on standby power, then an auxiliary station in a mobile unit, and

finally the trailer communications center will take over in case of failure of the first three.

In event of an alert, the operators will go to the trailer, whose location will be known to them at all times.

The operators will monitor the various services, including the New York State attack warning system, and coordinate alerting that remains to be done. All alerting and general instructions will be transmitted on the coordinating frequency, and received by the selective-dispatch receivers. Transmissions will be kept at a minimum, as will transmissions on all frequencies. The primary consideration will be to allow the various services to continue their own dispatching so long as they are able to do so in accordance with the established plan set forth by Civil Defense Authorities.

Following the dropping of bombs, or other disaster, the trailer center will be moved to a location near the scene, or to an advantageous site previously determined. It will be moved by one of the tractors offered by four trucking companies.

Much of the equipment for this system is currently being installed. The system is scheduled to be ready for operation by mid-August.

Remote Control of

Simple follow-up control circuit using two double-triodes permits moving industrial valves, levers and other motor loads to any desired position at full speed merely by changing the setting of a remotely located control knob

HERE are many applications for simple electronic controls that operate relays to make a motor turn at full speed in one direction or the other to move an object or a load into a desired position. A small motor (up to 1/8 hp) may operate directly from the relay contacts. For large motors to 1½ hp, these relays may control the contactors of a reversing switch.

The electronic positioning control circuit shown in Fig. 1 is a general-purpose follow-up control that may be used with any type of motor to move an object into any position selected by turning a small dial.

As an example, a motor-operated valve can be controlled from a distance by turning potentiometer slider A to any point on its dial; immediately, perhaps 500 feet distant, the motor will turn at full speed until it has brought potentiometer slider B on the valve to the position that matches the setting of slider A.

Briefly, when no action is desired and the slider of B is in the same position as the slider of A, transformer T receives no voltage so tubes V_{14} and V_{18} are both passing current, and neither V_{2A} nor V_{2B} is passing enough current to pick up its relay. If the slider of A is moved up, this turns off V_{14} so that V_{24} passes current and picks up relay 1; the motor then turns FORWARD and its shaft drives the slider of B up also. Conversely, if the slider of A is moved down, this turns off V_{1B} so that V_{2B} picks up relay 2; the motor then runs in REVERSE so that the slider of B is moved down also.

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Although a-c voltages are applied to the plates of all tubes, no tube can pass current during the negative half-cycles of plate voltage. During the positive half-cycles, the left-hand terminal of dead-zone potentiometer R is 4 volts more negative than the cathode of V_{14} . When transformer T is producing no voltage (so that no voltage appears across R_1 , R_2 , C_1 and C_2), the grid voltages of V_{1A} and V_{1B} may be varied between -4 and +14 volts, depending on the setting of R. Even when R is set so that the grids are negative, V_{14} and V_{18} pass a tiny current which, flowing through the high resistance of R_s or R_s , causes a large voltage drop across these resistors. The grids of V_{24} and V_{2B} are held far more negative than their cathodes, so tubes V_{2A} and V_{2B} do not pick up relays 1 or 2. This condition continues so long as the slider of A is not turned away from the same position as the slider of B.

Potentiometers A and B are connected to the same 115-volt a-c supply, with polarity such that when their sliders are in the same position, the two sliders are at the same potential and no voltage is applied to transformer T.

Follow-up Action

Assume the slider of R is set at its extreme left, for minimum voltage. As shown in Fig. 1C, the potential of the slider now is close to the cutoff potential of V_1 . If the slider of A now is turned a small amount to the right, a small voltage difference (often called the error voltage) is applied to transformer

T during the positive half-cycles to make its secondary terminal positive.

Position-controller potentiometer A causes a phase shift between the signal voltage of T and the line voltage (or anode voltage of the tubes). Capacitor C_1 acts with R_1 to bring the voltage across C_1 in phase; this is the voltage used to control V_{14} , as shown in Fig. 1B. This small signal voltage from T forces the grid of V_{14} closer to cutoff, reducing the plate current through R_3 and thereby swinging the grid of V_{2A} in a positive direction so plate current of V_{23} increases to pick up relay 1. The same signal voltage has the opposite effect on tubes V_{1B} and V_{2B} , hence relay 2 remains de-energized.

Tubes V_1 and V_2 actually serve as a two-stage amplifier. Moving the slider of A through half a degree causes only 1/5 volt across the secondary of T. This tiny voltage at the grid changes the plate current of V_{14} enough to vary the grid of V_{24} by 6 or 8 volts; this makes the plate current of V_{24} increase by 10 milliamperes, enough to pick up relay 1.

When relay 1 is picked up, a circuit is completed to the FORWARD contactor coil. The motor drives the valve load in the forward direction and also moves the slider of B to the right. This movement decreases the error voltage applied to T until the V_{14} current increases and relay 1 drops out.

If the slider of A now is turned to the left, a voltage difference (or error voltage) again is applied to transformer T, but with opposite phase relation to the previous error voltage. Now, as shown in Fig. 1C, the grid of $V_{1.4}$ becomes more positive and relay 1 is not affected; the

This article is based on material contained in a book by the author, "Electronic Motor and Welder Controls", published this year by McGraw-Hill Book Co.

Positioning Motors

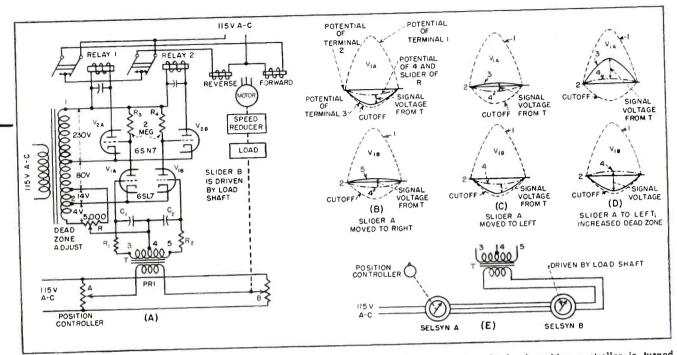


FIG. 1—General-purpose follow-up circuit for actuating contactors of reversible motor when knob of position controller is turned, waveforms of circuit voltages, and alternative controller arrangement using selsyns. Numbers on curves refer to potentials of correspondingly numbered transformer terminals in the circuit

grid of V_{1B} is forced more negative, decreasing the plate current of this tube and swinging the grid of V_{2B} more positive so that this tube picks up relay 2. This relay completes the circuit to the Reverse contactor coil, so the motor drives the valve load in the reverse direction and moves the slider of B to the left until the error voltage at T decreases and relay 2 drops out.

Accuracy Versus Dead Zone

When dead-zone adjusting potentiometer R in Fig. 1A is set for minimum potential, a very small movement of the slider of A (perhaps only 4 degree) may pick up relay 1 or 2; the motor turns, trying to position the load (such as a valve) and the slider of B to within this 1-deg accuracy. However, such close follow-up or accuracy can rarely be used unless the motor is stopped quickly (as by a brake on the motor shaft). When the motor brings the slider of B to the position where relay 1 drops out (thereby removing power from the motor), the motor and its load may coast far enough to move the slider of B beyond the slider of A so that relay 2 picks up, reversing the motor. This may cause the motor to turn back and forth, although the slider of A is not being moved; this hunting action must be stopped.

If potentiometer R is now adjusted to raise the potential of its slider, as shown in Fig. 1D, the slider of A must be turned a greater amount (such as 2 deg) before the signal voltage of T is large enough to force the grid of V_{1A} low enough to make V24 pick up relay 1. Similarly, as the motor drives the slider of B toward the same position as the slider of A, relay 1 may drop out while the slider of B still is 2 deg out of line. This permits the coasting motor to drive the load and the slider of B through an extra 2 or 3 deg without causing $V_{\scriptscriptstyle 2B}$ to pick up relay 2. Advancing the slider of R thus produces a larger neutral or dead zone within which no correction occurs.

If the motor and load coast far enough to move the slider of B through 3 deg after relay 1 drops out, then R must be set to provide a dead zone of nearly 4 deg; this limits the accuracy to plus or minus 2 deg, since the slider of A may be

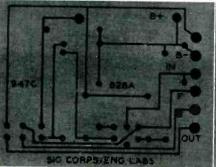
turned 2 deg before either relay operates. The accuracy of any follow-up system depends on the width of the neutral or dead zone needed to prevent hunting.

Selsyn Control of Positioning

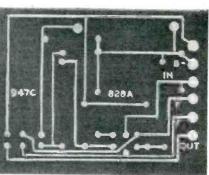
In place of potentiometers A and B in Fig. 1, two selsyns may be used as shown in Fig. 1E. Three wires connect the stationary parts of the two selsyns. The rotor of selsyn A may be turned through 180 deg by the position controller dial; the rotor of selsyn B is driven by the load shaft. When the rotor of B is exactly in line with the rotor of A, no voltage is delivered by selsyn B to transformer T, so the motor does not move the load. If selsyn A is turned to the right, selsyn B produces an a-c voltage at T that may turn off tube $V_{\scriptscriptstyle 14}$ and pick up relay 1; the motor then drives selsyn B to the right, decreasing this a-c voltage. When selsyn A is turned to the left, the voltage produced by selsyn B is of opposite phase and turns off V_{24} , picking up relay 2; the motor then drives selsyn B to the left until relay 2 drops out.



Copper-foil-coated phenolic chassis is starting point for auto-sembly of battery-powered audio amplifier



2 Conductor pattern is applied in acidresistant ink by silk-screening or offset printing



Chassis after etching away unwanted metal in ferric chloride solution and removing resist

Auto-Sembly of Miniature

THE SYSTEM of miniature circuit assembly to be described was evolved as a result of several Signal Corps study programs, aimed at providing relief from the production, performance and maintenance problems attendant on highly compact sub-miniature equipment designs. The system designated as auto-sembly borrows the very convenient prefabricated conductor pattern of printed circuits, and permits rapid and effective electrical combination of this pattern with conventional quality components by a one-shot solder dip process. The elements of auto-sembly are: (1) Formation of the conductor pattern; (2) selection of components; (3) rapid assembly: (4) packaging.

By S. F. DANKO and S. J. LANZALOTTI

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Of the many techniques of metallization extant today, the etching of copper-faced laminates (electrolytic copper foil bonded to an insulating base) provides one of the most convenient means of pattern fabrication. In this subtractive process, a resist image of the desired conductor pattern is printed or otherwise delineated on the copper foil surface and the laminate then exposed to an etchant. The resist material protects the metal against the action of the etchant

during the short etching period. The Signal Corps Laboratories demonstrated early in 1949, in collaboration with the Etched Products Corp. of Long Island, that the commercial offset presses of the etched nameplate industry could be readily adapted to the deposition of such resists on commercially available foil laminates for subsequent etching.

The photo-etch technique of the graphic arts industry was equally well adapted to such resist fabrications, particularly where higher orders of definitions or reproducibility were desired. Several other metallizing techniques, including the Mallory pressed powder technique and the Franklin Air-Loop stamping process, are also compat-

Mobilization Advantages

SAVES LABOR by eliminating manual wiring, which ordinarily requires more manhours than any other step in electronic equipment manufacture

MINIMIZES REJECTS because etched wiring is absolutely uniform from chassis to chassis

ELIMINATES UNSOLDERED JOINTS because all joints are made simultaneously in one solder-dip operation

REDUCES INSPECTION TIME because mechanization reduces probability of errors practically to zero

SPEEDS ASSEMBLY TIME because leads of components are simply pushed into holes, with no hook or wrap-around joints

PERMITS GREATER MINIATURIZATION, because etched wiring takes practically no space and lends itself to stacking of chassis layers

IMPROVES RELIABILITY, because auto-sembly practically eliminates short-circuits between connections and leads

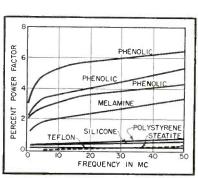
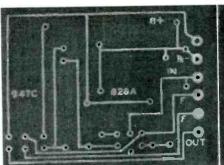
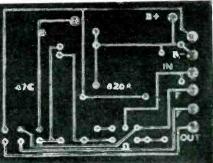


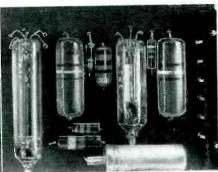
FIG. 1—Variation of power factor with frequency for some of the double-faced laminates used in auto-sembly. Note variations in phenolics from three different suppliers. Melamine, polystyrene, silicone and Teflon materials are glassfabric or glass-mat loaded



4 Next step is punching or drilling of holes through which leads of components are inserted from rear



5 Underside of chassis after components are mounted and assembled by one-shot solder dip



Top of completed chassis, showing layout of tubes and components, with plug-in terminals at right

Military Equipment

Production of wiring by acid etching of copper-faced insulating laminates permits rapid assembly of high-quality conventional JAN components by inserting leads in punched holes, then immersing chassis surface in molten solder to make all joints simultaneously

ible with the auto-sembly system since they yield high-conductivity patterns which are amenable to component assembly by solder dipping.

It is the intention of this presentation to emphasize the etched-foil technique of preparing a conductor pattern, inasmuch as this process lends itself excellently to pattern fabrication by the laboratory designer for prototype model constructions. The choice of a process for forming the pattern for production depends on such factors as the detail required in the pattern, quantities involved, tooling costs, production rate and delivery time.

Laminates

The physical characteristics of the commercially available copperfaced laminates which are of prime interest for auto-sembly include bond strength of the foil to the laminate, compatibility of this bond with the temperature shock conditions which it will experience during solder dipping, dielectric constant, power factor, are resistance and moisture absorption. Table I lists in a general way some of this physical data for comparison purposes, and Fig. 1 shows

the relation of losses to frequency for several of these laminates. It should be noted that these loss curves apply where the base material is used as the total dielectric, between metallized faces on opposite sides of the material. In the usual application, where capacitances between adjacent lines are of concern, the losses are substantially lower than indicated since the dielectric is part air.

The laminates are also available

with copper on both sides to permit fabrication of aligned patterns on both faces. Most common thicknesses of copper are 0.00135", 0.003" and 0.005".

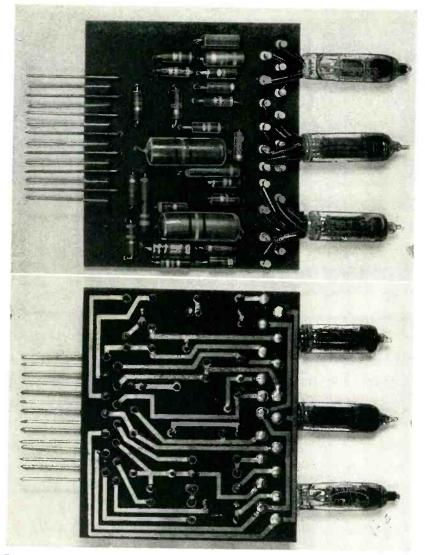
Resists

Within the design laboratory, screening of the resist is perhaps the simplest and most convenient method of printing the pattern prior to etching. The screens themselves may be prepared by any of

Table I—Characteristics of Copper-Faced Laminates

Laminate Chassis	Reported Bond Strengths* (1b)	App. Dielectric Constant	Power Factor at 1 mc	Mois- ture Ab- sorption (24 hr)	Arc Resist- ance	Max. Oper- ating Temp.
XXXP Phenolic	6-8	3.5-5.5	2.8 %	1.3 %	Poor	125°C
Melamine— Glass Fabric	7-9	6.8	1.3 %	2.0 %	V G	135°C
Silicone— Glass Fabric	2-8		0.35%	0.3 %	V G	200°C
Teflon— Glass Fabric	3-7	3.3	0.10%	0.3 %	V G	200°C 85°C
Polystyrene— Glass Mat	2-3		0.38%	0.55%		05 C
Polyester— Glass Fabric	(Under Dev	ler Development) Fair			

^{*}Bond strengths are measured by pulling a scored one-inch copper strip vertically away from the laminate surface. The bond strengths vary among manufacturers but the values shown indicate the general range of values measured. (The bond strength of copper-phenolic is an exception, being representative of one manufacturer's item.)



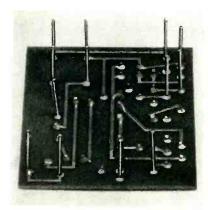
Top and bottom views of auto-sembled chassis containing phase-shift oscillator, avc, and clipper-limiter circuits that serve as a unitized packaged plug-in assembly for a complex electronic fault locator. Embedment of chassis in casting resin will provide protection against moisture and give rigidity to plug-in terminals at left

the many commercial screening organizations or may be prepared within the laboratory without too much art or effort. For most general line work for conductors, a screen mesh of about 150 per inch is adequate.

Satisfactory resists for screening are many, but synthetic screening enamels available from the art supply houses have been particularly useful in this application. An example is the Synthetic Screen Enamel, series 1500, any color, obtainable from the Colonial Process Supply Co., New York City. Resists made by dissolving asphaltum in an appropriate vehicle are also convenient, but care must be taken to get the proper screening consistency. The synthetic enamel resists require approximately 8 hours to

dry (overnight, or may be oven dried in a few hours).

The foil laminates with the resist patterns may be etched in a rocking tray or may be mounted copper face down about one-half inch over the surface of the etchant, which can then be agitated by bubbling air violently through the solution. The recommended etchant for copper is a 50-percent solution of ferric chloride. This will etch through a 1.35-mil foil in about 5 to 7 minutes depending on the freshness and temperature of the bath. The patterns are then washed in running water, following which the enamel resist can be removed by use of a thinner or by light abrasion. Commercially these patterns would then be processed through a die stamping operation to provide the neces-



Detailed view of solder-dipped pattern.
Fillets at termination points are obtained by using a waxy flux. Long protruding leads go between components and into holes in next layer

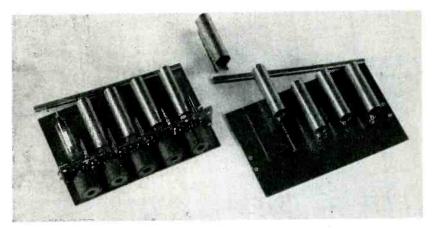
sary holes through the pattern at each termination point. In the laboratory, of course, these perforations would be drilled or punched.

Conductor Characteristics

The thinner copper foils have more than adequate conductivity for general miniature circuit applications, as indicated in Table II. A 1.35-mil-thick conductor 32-inch wide on a representative 3 x 2-inch phenolic plate will carry over 3.5 amperes continuously with a 40-degree C maximum temperature rise of the conductor. A similar 3-milthick conductor under the same conditions will carry over 5 amperes. A studied effort has been made to discourage use of such ratings as fusing currents, peak current capacity and similar shorttime ratings which do not provide useful information for a designer's

For general conductor applications a line width of about 32-inch is recommended, although widths down to 1/64-inch are dipped consistently. Experience to date indicates that an area of metal equivalent to about a 36-inch-diameter circle at each junction point is desirable since a well-formed and substantial fillet of solder will then be built up. Spacing of lines, in general, should be kept at least at 32-inch unless otherwise demanded by pattern layout.

On phenolic surfaces, the peak flash-over voltages at room ambient conditions with no coating protection over the 32-inch pattern spacings were about 1,800 v. This



Experimental five-stage impedance-coupled i-f amplifier having heat-dissipating metal back plate and individual metal tube envelopes for conducting tube heat to the plate

s'z-inch spacing has been used consistently on 300-v d-c circuits, with no failures recorded. In this respect, the desirable arc resistance features of some of the laminates are strong points in their favor. With this spacing, insulation resistances well over 100,000 megohms under ambient room conditions are measured between 3-inchlong copper lines on phenolic with no protective finishes applied.

The distributed capacitance between lines is significant and cannot be ignored at the higher frequencies. On XXXP phenolic, this capacitance is of the order of $1 \mu \mu f$ per inch for 32-inch spacing and 0.7 uuf per inch for 1-inch spacing. These values will be lower for the lower dielectric constant materials. Proper pattern layout will allow use of very short sections of r-f conductors, and it has been convenient at times to run short ground lines between two critical conductors to reduce this electrostatic coupling to a still lower figure.

Although the inherent power factor of such materials as phenolic is fairly high, the effective power factor is considerably better inasmuch as the dielectric of the distributed capacitance is part air and part insulating material. As a generality, this pattern capacitance can be said to be comparable in magnitude to that which may be encountered in conventionally wired compact assemblies, but with the advantageous feature of being identical from pattern to pattern.

The commercial offset press process can yield line work of the order

of 8 to 10 mils in width as a matter of routine production. The photoetch technique yields conductor widths down to about 4 mils without premium charges. These higher definitions are useful in delineating the precise geometry required of spiral-formed coils such as have been used in television tuners and high-frequency i-f coil patterns.

It is recommended that the finished copper patterns be silver-surfaced by dip plating (which yields a satisfactory flash coat) or by direct plating to eliminate oxidization of the metal on long storage with the subsequently increased difficulties of cleaning. The commercial etching houses will supply such surfacing as a matter of normal production processing.

Table II — Resistances of Etched Copper-Foil Conductors

Line Width	Ohms per Inch at 20°C for 100% Conductivity Copper					
(inches)	0.00135"	0.003"	0.005"			
	Thick	Thick	Thick			
1/100	0.050	0.023	0.014			
1/64	0.032	0.015	0.0087			
1/32	0.016	0.007	0.0043			
3/64	0.011	0.005	0.0029			
1/16	0.008	0.0035	0.0022			
3/32	0.0054	0.0024	0.0015			
1/10	0.0050	0.0022	0.0014			
1/8	0.0040	0.0018	0.0011			
5/32	0.0032	0.0015	0.0009			
3/16	0.0027	0.0012	0.0007			
1/4	0.0020	0.0009	0.00055			
3/8	0.0013	0:0006	0.00037			

Copper has been mentioned exclusively as the conductor material because of commercial availability of the copper-surfaced laminate. However, some of the etching sources which do their own laminating have prepared patterns of silver, iron, aluminum, brass. bronze and even spring-tempered steel for special purposes. Some of these metals have proven particularly useful in delineating patterns for switches. It should be mentioned, too, that these laminatorsetchers can supply embossed, flush or subsurfaced patterns for special purposes or provide upset metal nodules on the patterns where required for switch contact points.

Components

Auto-sembly requires that the components have stiff protruding terminations, preferably of a radial type, such that the component can be conveniently inserted into its designated place on the insulated side of the chassis. Resistors, capacitors or other components having axial leads can have these leads cut and shaped by jigging machines built for this purpose. The production arrangement can be such as to have the operator hold the conventional component momentarily in the jig for shaping and then in the next motion insert it into place on the pattern. The alternate proposition would be to supply the operator with pre-jigged Components where components. stiff protruding terminations are desired include termination devices such as sockets, connectors and receptacles. Current Signal Corps developments aim to provide such electro-mechanical devices that are particularly adapted to auto-sembly or other card-type systems.

Most conventional components can be readily adapted to auto-sembly mountings. The use of conventional components is considered a desirable and realistic approach in this miniaturization technique. The designer has available for his choice not only the variety of components but also the full range of tolerances, electrical characteristics and physical sizes in the established quality lines of JAN or commercial components. There are no special skills imposed on the assem-

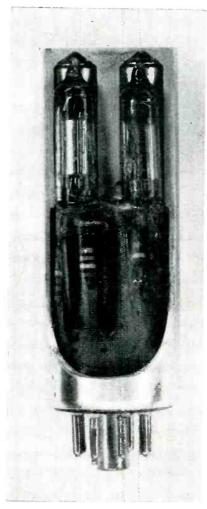
bler of the equipment such as would be required if he were to deposit or print his own resistors or capacitors. The fabrication of such deposited components is still a highly specialized art. Where economic or performance advantages can be realized in any fabrication through the use of these printed circuit aggregates, then such groupings of deposited components (such as Couplates, Bulplates and multivalue capacitors) can be treated like any other component for autosembly.

Assembly

The use of the one-shot solderdip technique has been particularly convenient in the fabrication of highly miniaturized assemblies. In this operation the insulating base, with the fluxed pattern on the underside and the components mounted on the topside, is momentarily dipped in a eutectic solder bath (63 percent tin, 37 percent lead) maintained at about 450 F. The period of dip recommended is of the order of two or three seconds, but this time may be increased if large areas of copper are left on the pattern for shield purposes or as heat radiation surfaces. Periods of dip as high as ten seconds have been made without delamination or deleterious effects on such base materials as phenolic, Teflon or silicone.

The success of the solder dip procedure is particularly dependent on the flux used just prior to solder dipping. The following formulation has been found successful: 1 part Glyco Wax No. S932 made by the Glyco Corp., Brooklyn, N.Y.; 1 part Kester No. 1015 (activated rosin in alcohol flux) made by the Kester Solder Co., Newark, N. J.; 1 part toluene (more toluene may be added if a thinner consistency is desired). The bath should be kept warm during use (about 110 F). The flux itself may be applied by dipping the pattern side momentarily in the flux or by brushing the flux over the pattern.

It is important that the pattern be removed from the surface of the solder bath with a sidewise lifting motion. This technique yields a soldered pattern with well-formed fillets at each termination point and a more or less hemispherical cross-



One-watt audio amplifier with portion of metal heat-dissipating container cut away to show use of folded circuit. Components are auto-sembled on flexible Teflon chassis that is relled up and inserted in container, then cast in resin. Subminiature tubes fit into machined holes in solid metal end of container

section of solder on the conductor lines. The pick-up of solder on the conductor lines is extraneous and incidental to the process but it does add to the total conductivity of the conductors. One manufacturer on a production line basis has evolved a variation of this technique wherein his solder dip procedure deposits mechanized solder only at the termination points, a desirable process from the point of view of solder economy.

Packaging

Packaging includes ruggedization, moisture protection, adequate heat transfer and provision of connective components such as may be required for plug-in use. Use of right-angled connectors provides a convenient means of integrating a connector quickly to the pattern since such a connector is treated as any other component in the course of the assembly operation.

Although stiff laminates have been discussed up to now, flexible laminates have been used as the chassis for the pattern in making folding circuits. Assemblies on thin sheet Teflon glass laminates and on silicone rubber (thicknesses of 3 to 10 mils) have been autosembled and then folded and inserted into capsules.

Conclusions

The complete and detailed coverage of all aspects of the autosembly system have not been presented here because of the voluminous nature of an exhaustive treatment of the subject. It can be shown, for example, that sections of auto-sembly circuitry can be decked by providing completed solder-dipped auto-sembled sections with stiff swaged-on protruding connectors which can be inserted into place on another deck and then solder-dipped into place. Crossovers can be made by eyeletting and subsequent spot-soldering on double-faced laminates, or by use of staple-shaped wires as compo-

The information as presented has been intended to show some of the more basic concepts and techniques of auto-sembly and the major considerations which govern its current applications in service and commercial equipment application. The use of the etched pattern and the auto-sembly system of integration both provide effective tools which are not necessarily limited to electronic circuit fabrication. These patterns, for example, are used in the preparation of heaters, Faraday shields, switching devices, and in one instance even for forming low value bypass capacitors. Autosembly is being used in new military equipment designs and is currently applied commercially to repetitive types of circuits such as are used in counters. The extrapolation of auto-sembly to other applications is dependent almost entirely on the imagination of designers familiar with the capabilities of these tools.

Automatic TV Sync Lock

Facilities for special effects can be applied to an incoming video signal from a remote point, such as a network broadcast. The local sync generator is synchronized to the remote with horizontal afc control and phase-correction control of the vertical interval

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Television programming requirements continually demand increased versatility on the part of the technical facilities.

Equipment is available to perform various special effects such as montage, superposition, lap, fade and wipe dissolve, as long as all program material is obtained from studio and film channels referenced to the same synchronizing generator. The application of these local technical facilities to program material originating at a remote point necessitates that the local synchronizing generator be synchronized, both in frequency and phase, with the mixed sync of the composite remote signal. A sync lock unit performs this synchronizing function.

With a sync lock system a picture signal coming in from a remote point of origination, such as the signal from a network, an off-the-air receiver, a satellite studio, or a remote pickup, can be treated and handled just as though it originated in an adjoining local studio. All of the special effects facilities can be applied to this remote signal.

For example, local commercials can be surreptitiously inserted into a network show of a nationally advertised product. On a local remote pickup it is not necessary to transport extensive additional equipment to the pickup point or, alternatively, to switch composite video signal to the local sync generator, with a flapping disturbance, for each commercial or local program insertion.

A remote signal, either from the network or off the air, frequently exhibits bad spiking of the front porch, sync distortion, obliteration of the sync by noise or improper front-porch width. With an adequate sync lock system such a distorted composite remote signal can be reconstructed with clean sync.

The sync lock must provide solid and undisturbed synchronization of the local sync generator. Additionally, it must provide front-porch width adjustment, remotely controlled from the station video control position, so that the time phase between remote signal blanking and reintroduced local sync can be matched quickly to the local system setup. For maximum utility the sync lock should be automatic in

operation and free from critical adjustment such as clipping level controls. A sync lock unit which meets these requirements has been developed.

Locking Method

Synchronization of the associated local sync generator to the remote sync generator is accomplished by automatic frequency control of the master oscillator of the local sync generator and automatic phase-correction control of the vertical interval. Since the remote signal is composite video and sync, as obtained from a network connection or from an off-the-air receiver, it is necessary to separate out the video and retain the

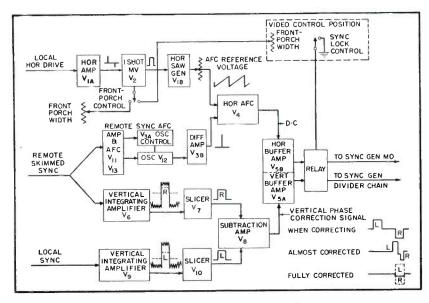


FIG. 1—Stages of the sync lock and their functions

skimmed sync for use as the input remote mixed sync signal. Skimming of the sync is normally accomplished through the use of a stabilizing amplifier.

The basic functions of the various portions of the sync lock are shown in Fig. 1. There are only two operational controls, both of which are most conveniently located at the video control position of the station. These controls are the sync lock REMOTE-LOCAL with push-button switch and the FRONT-PORCH WIDTH adjustment for the sync lock.

Horizontal AFC

The horizontal afc, V_{**} , performs the function of automatic frequency control of the master oscillator of the local sync generator. It utilizes a sawtooth waveform, derived from the local horizontal drive, and a narrow pip derived from the horizontal sync of the incoming remote sync signal.

The narrow pip allows the instantaneous value of the sawtooth voltage occurring at the time of the pip to charge the output capacitor of the afc. This voltage controls the frequency of the master oscillator of the sync generator. If this oscillator, and hence the repetition

rate of the sawtooth, is not at the correct frequency the subsequent pips will alter the d-c output voltage of the afc, thus changing the oscillator frequency.

The phase relations of the afc, the d-c control voltage and the oscillator frequency are such as to adjust the frequency so that the pip will always occur at the same voltage level point on the fast part of the sawtooth. Any attempt on the part of the oscillator to change frequency is immediately counteracted by a change in the d-c controlling voltage in a direction to correct the frequency change.

The narrow pips at horizontal rate, to feed the horizontal afc, are derived from the incoming remote skimmed sync. They are maintained synchronous, even though the input sync may be badly distorted by noise, by the utilization of a closed-loop sync afc system. These circuits are analogous to the horizontal sync circuits used in a modern receiver. In this adaptation the sync afc is a pair of diodes, a part of V_{11} and associated circuits.

Filter circuits are employed to stabilize the afc and to prevent frequency modulation of the generated narrow pip, output of V_{ss} , caused by the disturbing action of

the equalizing pulses and serrated pulses occurring during the vertical interval. The d-c output of the sync afc controls the effective resistance of the oscillator control tube, V_{3a} . Change in this effective resistance controls the frequency of the syn oscillator, V_{12} .

Although this is a sine-wave oscillator, the circuit arrangement is such as to generate a sawtooth waveshape at the plate and this is fed back to the sync afc. Also, the sawtooth is differentiated in the grid of V_{sB} and amplified, thus forming the narrow pip signal required to feed the horizontal afc.

The sawtooth waveshape for the afc is generated in V_{1B} and is synchronous with the local sync generator horizontal drive. The d-c axis for this a-c voltage is set by the afc reference voltage, the value of which is determined by the particular sync generator with which the sync lock is associated.

The time duration of the fast part of the sawtooth is determined by the width of the pulse generated in the one-shot multivibrator, V_2 , and is adjustable over a suitable range by the front-porch width control.

Front-Porch Width

The multivibrator pulse, the length of which determines the front-porch width, is initiated by the leading edge of the local horizontal drive input signal. Since the width of the multivibrator pulse is controlled by a d-c voltage, a very simple remote control of the front-porch width is obtained.

The way in which an incorrect front-porch width on the incoming remote sync is corrected through the front-porch width adjustment, where the system is operated with reintroduced local sync on remote blanking, is shown in the phase relations diagrams of Fig. 2.

Figure 2 shows the conditions when the remote sync has a narrow front-porch width. For illustration, remote blanking, sync, and front-porch are assumed as 11, 4.8, and 0.5 microseconds respectively. The narrow pip signal, output of $V_{\scriptscriptstyle 3B}$, is shown as being one microsecond delayed from the leading edge of remote sync. This delay of about one microsecond was designed into the

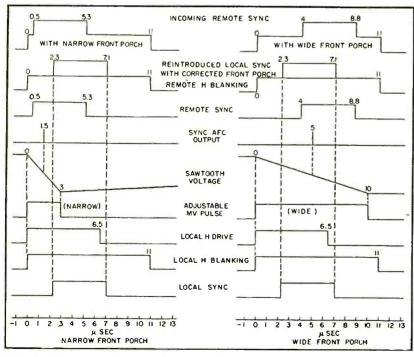


FIG. 2—Time-phase relations when correcting narrow and wide front porch on incoming remote sync

sync afc to obtain the most effective range of front-porch width adjustment.

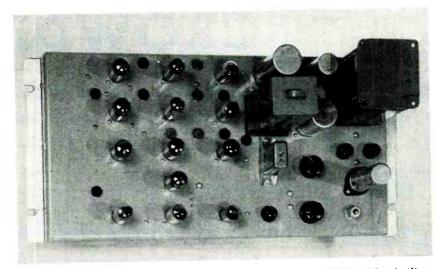
If the afc reference voltage has been centered with respect to the local sync generator master oscillator control voltage, the sawtooth voltage, output of V_{1B} , will straddle the narrow pip signal equally in time as shown. This fast part of the sawtooth was caused by the multivibrator pulse which, in this example, was set to be 3 microseconds in length. Since the leading edge of local horizontal drive initiated the multivibrator pulse, and since local sync, as determined in the sync generator, is assumed as 2.3 microseconds later, reintroduced local sync will time-phase with remote horizontal blanking as a front-porch width of 2.3 microseconds.

When the incoming remote sync has a wide front porch, correction is obtained by setting the frontporch width adjustment so that the multivibrator generates a long pulse, as also shown in Fig. 2. In this illustration the front-porch width on the incoming remote sync is assumed as 4 microseconds. By setting the length of the pulse to 10 microseconds the time phase of local horizontal drive and local sync will be shifted so that reintroduced local sync will again be corrected Thus the to 2.3 microseconds. width of the multivibrator pulse serves to alter the time phase of the local sync generator relative to the remote sync and blanking, hence providing front-porch width control.

Vertical Phase Correction

Automatic phase correction of the vertical interval is obtained by the comparison in time phase of pulses corresponding to the serrated intervals of the local and remote sync. When these pulses are not in time coincidence, that corresponding to the local is used to suspend or hold the count down between the twice-horizontal rate pulses and the counted-down field rate pulse of the local sync generator by, presumably, four half-lines each field. This suspension of count continues in subsequent fields.

As the correction approaches finality, the pulses from the remote



Rack-mounting chassis connects to associated equipment by coaxial cables feeding from backside

and local start to overlap. A portion of the positive-going local pulse disappears in the negative-going remote pulse, thus reducing the number of suspended counts. When the phase is fully corrected to coincide with the correct field, the correcting signal has been cancelled. Since at the time of initiating sync lock the vertical phases of the local and remote are perfectly random, it works out that the most probable value for the time of vertical pull-in is 1.1 seconds. Horizontal pull-in is practically instantaneous, a matter of only a few horizontal lines.

In obtaining the vertical rate pulses corresponding to the serrated intervals, the mixed sync in both local and remote channels is integrated and then sliced. The slicing operation is equivalent to a simultaneous top and bottom clipping and is accomplished with circuits which require no clipping level controls, nor are any provided. Slicing is precise and independent of the channel input voltage between the design levels of 3.5 to 8 volts.

Slicing on the local channel is done at a higher level on the integrated pulse than on the remote channel, thus assuring that the local pulse will always be slightly narrower than the remote pulse so that, in the subtraction amplifier, the positive-going local pulse will always completely disappear in the negative-going remote pulse when phase coincidence is obtained.

The General Electric Sync Lock Type TV-30-A was designed to operate in conjunction with synchronizing generators of both GE and RCA manufacture. Operation with other sync generators has not been fully investigated.

System Considerations

In system operation with a sync lock, especially when a station has only one sync generator, consideration should be given to the time required for the local sync generator to get back to local power line lock when it is desired to leave the This situation remote signal. should be anticipated at the termination of a remote program since network sync may be opened as a result of repatching of the system. The time requirement is dependent upon the design of the sync generator and has nothing to do with the sync lock. Sync generator GE type PG-2-B, which employs a comparatively rapid afc, is especially adaptable to a sync lock system since pull-in time to local power line lock can normally be accomplished during the switching time from remote to local picture.

Consideration also should be given to the local film projector to obtain maximum flexibility and utilization of a sync lock system. When the special effects are to be applied between a remote program and local movie film, it is necessary that the projector be completely synchronized to the local sync generator rather than to the power line. Such projectors are now available thereby making possible a completely synchronized system.

CROWDED-BAND

Transmitter-receiver combination operates in 148 to 174-mc range. Receiver uses tripletuned i-f transformers to minimize adjacent-channel interference and controlled r-f gain to minimize intermodulation interference. Field test results are given

HE ever-increasing use of the channels in the 148 to 174-mc range has brought with it many equipment design problems. The solution of these problems has led to the development of many improvements in receivers and transmitters. The design of equipment capable of performing satisfactorily under present-day crowded-band conditions is further complicated by the requirements of simplicity, reliability and ease of serviceability.

Types of Interference

There are two general types of interference that are especially troublesome to mobile equipment operating in this range. These are adjacent-channel interference and intermodulation interference. These two types are quite different, both in their nature and in the measures that must be taken to minimize them.

Adjacent-channel interference caused by the presence of a strong signal on the next channel 60 kc off, takes three forms, desensitization or reduction of gain due to self-bias being developed on the grids of the i-f amplifiers or mixers;

break-through of adjacent channel signal reaching the discriminator although greatly attenuated; and noise interference produced by onfrequency noise sidebands radiated by the adjacent channel transmitter.

Desensitization is minimized by keeping the gain ahead of the selectivity determining stages (usually low i-f stages) as low as possible consistent with good sensitivity. With good design desensitization effects can be kept small compared with noise interference.

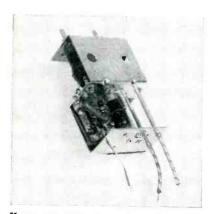
Modulation break-through can be reduced until it is less than the noise interference by adequate i-f selectivity. Noise interference, which is a characteristic of the transmitter and, therefore, unaffected by receiver design, will then be the limiting factor in adjacentchannel interference. A study of the various ways of achieving the required selectivity showed that triple-tuned low i-f stages in a double superheterodyne circuit resulted in the most efficient use of tuned circuits, the fewest number of tubes, and the simplest circuitry.

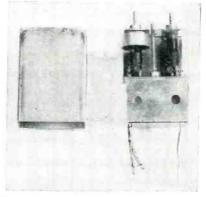
Intermodulation interference re-

sults from the generation of a carrier by two signals spaced n channels and 2n channels from the signal frequency, where n is any integer. This kind of interference results from components generated in the r-f tubes and appears as an on-frequency high i-f signal. Any circuitry after the first mixer will have equal effect on the desired signal as well as the interference.

The steps that can be taken to change the intermodulation characteristics of a receiver are limited. particularly with regard to that produced by the adjacent and the alternate channel signals. choke channel spacings these signals pass through any practical 150mc circuits virtually unattenuated. It takes a resonant circuit with a Q of the order of 2,000 to 3,000 to produce any appreciable attenua-This can be accomplished with a well-designed cavity resonator but it is impractical as the cavity is larger than the whole mobile unit and costly to build. Some improvement in intermodulation attenuation can be realized by using only enough r-f gain to obtain the sensitivity, required because changes in r-f gain will have greater effect on the intermodulation product than on the desired signal. It is more effective to reduce the gain of the first r-f stage than that of the second. For areas in which the desired signal is greater than the level required to produce 20-db quieting, it is advantageous to use automatic gain control to reduce the r-f gain.

In any receiver it is important to have enough selectivity to insure that the adjacent-channel performance is at least equal to the intermodulation performance. Freedom from intermodulation interference





Upper compartment of i-f transformer assembly houses primary and tertiary windings.

Secondary winding is housed in lower compartment

MOBILE EQUIPMENT

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at a given point is valueless if adjacent-channel breakthrough occurs. In a practical design, it is often advantageous for the adjacent-channel performance to exceed the intermodulation performance, for interference from the latter requires simultaneous operation of both interfering stations, and therefore in many applications will occur only a fraction of the time the adjacent-channel station is on the air.

Receiver Design

The receiver whose block diagram is shown in Fig. 1 was designed in accordance with the considerations mentioned above. It has been thoroughly field tested to prove its ability to operate under conditions of severe adjacent-channel and intermodulation interference resulting from the presence of both adjacent and alternate channels.

The receiver has a guaranteed 20-db quieting sensitivity of better than 0.6 microvolt and an image and spurious response rejection of greater than 95 db. In production the sensitivity is actually running between 0.35 and 0.5 microvolt. This is achieved by using five slugtuned coils of 1½ turns of bus wire, 1 inch in inside diameter, mounted in individual 3-inch diameter shield cans. Two ceramic tuning capacitors in parallel are mounted inside the can, so connected that by clipping an external link on the coil base the range can be changed from 148 to 162 mc to 162 to 174 mc. One of these coils has a capacitance tap to match down to a 50-ohm coaxial cable and constitutes the antenna transformer. The other four are coupled together in pairs with high side capacitance coupling to form two double-tuned r-f stages.

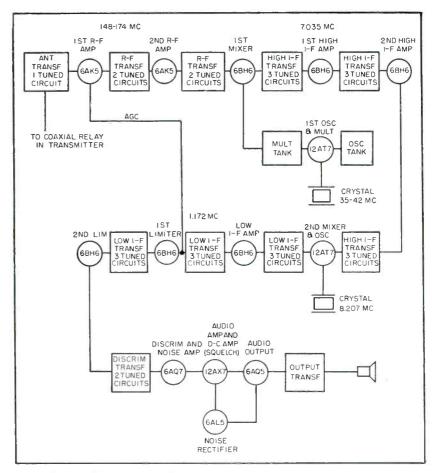


FIG. 1—Receiver block diagram shows double superheterodyne circuit

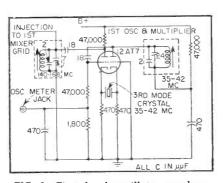


FIG. 2—First local oscillator employs third-mode crystal in 40-mc region

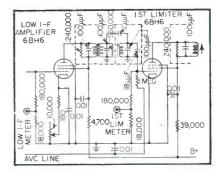


FIG. 3—Circuit of low i-f amplifier and first limiter shows use of triple-tuned i-f's

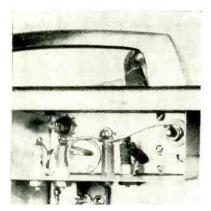
The performance of these tuned circuits compares very favorably with that of resonant concentric lines and this construction is much more simple and trouble free.

For the same considerations of sensitivity and r-f selectivity, 6AK5 tubes are used in the r-f stages, since the input resistance of any other suitable tube at 160 mc is much lower, resulting in increased

loading and loss of gain and selectivity.

Other than the individual coil shields, the only r-f shielding necessary is a vertical shield between the r-f stages and oscillator-mixer stage.

Another factor in the excellent spurious rejection of this receiver is the use of separate crystals for high and low i-f injection, permit-



Under-chassis view of low-pass filter used in series with antenna coil

ting the use of a third mechanical mode crystal for the high injection. This type of crystal operates at a frequency of from 35 to 42 mc depending on the signal frequency, and requires only quadrupling before injection. This greatly reduces the number of undesired oscillator harmonics that would be produced if a lower frequency crystal were used and the number of multiplications increased. Moreover, these fewer harmonics are widely spaced from the desired injection frequency, 35 to 42 mc for the nearest, so that the r-f selectivity gives ample protection.

A cathode-coupled oscillator circuit is used for the first oscillator as this circuit gives very stable operation with a mode crystal, and also provides multiplication by tuning the second tank circuit to a harmonic. The oscillator circuit is shown in Fig. 2.

Triple-tuned i-f transformers are used in the two high i-f stages. While the selectivity of the high i-f contributes only a little to the overall selectivity, it does protect the receiver against spurious responses occurring at the second mixer.

I-F Design

Both high and low i-f transformers are of the same basic design. A folded metal frame provides a means of mounting the coils and complete shielding between the primary and secondary. The primary and tertiary in the upper compartment are inductively coupled, the coupling being controlled by a coaxial shield around the upper part of the primary wind-

ing. The tertiary is coupled to the secondary by means of a feedthrough in the frame and a coupling capacitor. Connection is made to primary through two leads passing through plated copper tubing. Since the tubing is drawn down on the leads before assembly and the tubing is soldered to the frame, the leads are permanently anchored and there is no need for terminal boards to secure their upper ends. All the coils are tuned from the top. This i-f structure provides a compact, rigid transformer that is inexpensive to build. The windings are given a double impregnation with electrical grade resin lacquer and a wax flash dip to insure stability under the extremes of temperature and humidity experienced in operation.

A 12AT7 is used as a second mixer and second oscillator. The oscillator circuit has a ceramic trimmer that provides for small adjustments of second oscillator frequency so that the receiver can be zeroed on the transmitter with which it is to operate, without realignment of the discriminator and low i-f transformers.

The low i-f amplifier has a gain potentiometer in its cathode circuit. This permits the second limiter current for noise to be set initially to the value for optimum squelch circuit action and the gain to be brought back to this level as the gain decreases over a period of time due to tube aging. This feature should prevent unnecessary tube replacements and peaking of transformers. As long as the serviceman can maintain the proper second limiter current, he will know that he does not need to touch the tubes or transformer tuning.

A differential squelch circuit is used. This type of circuit is used to give a fast transition from

squelch to no squelch. A gain control in the cathode circuit of the noise amplifier tube allows the squelch to be set to the correct level for the noise conditions that are being encountered.

Transmitter Design

The transmitter block diagram is shown in Fig 4.

With the crowded condition of this portion of the spectrum, it is necessary to limit the swing of the transmitter to ±15 kc to prevent adjacent-channel interference. To do this, it is necessary to limit the slope, or rate of change, of the audio modulating wave. The modulation limiter employed to accomplish this slope limiting is composed of a pre-emphasis circuit, a voltage limiter, and a de-emphasis circuit, as shown in the block diagram.

The pre-emphasis circuit changes the audio input voltage into a wave whose amplitude is proportional to the slope of the audio input wave. The limiter tube is a double triode which amplifies in the first section and limits, as an overdriven amplifier, in the second section. A potentiometer in the output of this stage adjusts the point at which modulation limiting takes place. This symmetrically-limited output is fed to the de-emphasis circuit which changes the audio signal back to a wave whose amplitude is inversely proportional to the slope of the wave at the output of the limiter tube, and then to the grid of the modulator tube.

Thus, below limiting, the voltage output of the de-emphasis circuit is unchanged from the original audio input, while above limiting, the only change is that all slopes are limited to values that will prevent frequency deviation greater than desired. A visual representation of this limiting is shown in Fig. 5.

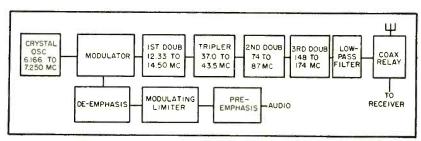


FIG. 4—Transmitter block diagram. Deviation at modulator output is \pm 625 cps

The modified Pierce oscillator is capacitance coupled to the grid of the modulator tube and inductively coupled to the plate of this tube. The output of this stage is then fed to the grid of the first multiplier tube as phase modulated r-f with a maximum deviation of ±625 cps.

As can be seen in Fig. 4, the frequency is then multiplied $2 \times 3 \times 2 \times 2$ or 24 times to the output frequency, and hence the deviation is $24 \times \pm 625$ or $\pm 15,000$ cps. Double-tuned coupling circuits are used from the tripler on to attenuate the spurious emission from the transmitter below the carrier frequency at least 70 db below the carrier output level.

The antenna circuit, which consists of a movable coupling coil with a series tuning capacitor, permits extremely simple tuning because the functions of antenna tuning and coupling are separate. The transmitter may be tuned into an an-

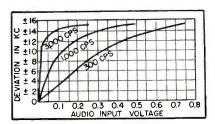


FIG. 5 — Typical modulation limiting curves for transmitter

tenna with a standing-wave ratio of 2 to 1 by merely resonating the final tank capacitor, resonating the antenna capacitor, and coupling to full power.

Field Tests

Two series of field tests were used to evaluate this equipment.

The first was conducted under conditions that might be frequently encountered in practice. It was made by using the transmissions of some of the City of Syracuse taxicab companies. Three of the companies operating on channels 5, 6, 7, are located as shown in Fig. 6A on level terrain. All of the transmitters have 50-watt output. They are equipped with high-gain antennas with gains of 2, 3 and 3, respectively at heights of approximately 105, 110, 115 feet, respectively. The receiver was tuned to channel 7 and

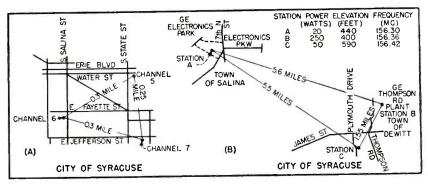


FIG. 6—Two series of field tests were made over area shown in these maps of the Syracuse area

the test car driven around the downtown area in the vicinity of the transmitter on channel 6. A number of test runs were made, and in no case was any adjacent-channel interference experienced even when the test car was in the street alongside the building upon which the channel 6 transmitter was located.

Careful checking revealed a few very restricted spots in the area around the channel 6 antenna where intermodulation interference could be heard when channels 5 and 6 were on but channel 7 was off. In every case, channel 7 completely captured as there was no trace of interference.

The second set of tests was made using special test stations on experimental frequencies.

Station A, a 20-watt station, was located at the Electronics Park Plant, elevation 406 feet, with a dipole antenna mounted on top of one of the buildings at a height Station B at the of 50 feet. Thompson Road Plant at an elevation of 400 feet, consisted of a 250-watt transmitter with a dipole antenna mounted on the roof at a height of 75 feet. Station C was a 50-watt transmitter, located on rising ground at an elevation of 590 feet, 12 miles from Station B. A high-gain antenna on a 38-foot mast gave station C an output equivalent to a 175-watt transmitter, and a relative height of 228 feet with respect to B.

The stations were operated on the following frequencies: A, 156.30 mc; B, 156.36 mc; C, 156.42 mc. The output of Station A was reduced to give a signal of 2 to 5 microvolts in the region of Station B, $5\frac{1}{2}$ miles away (see Fig. 6B), and tests of adjacent-channel and

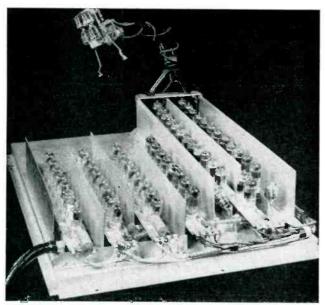
intermodulation interference were made in the vicinity of Station B with a receiver tuned to Station A. These conditions were far more severe than any usually encountered in practice but were so chosen that changes in design that caused only small changes in test bench measurements could be evaluated.

Under these conditions while slight adjacent channel interference was apparent in the background of the desired signal at a distance of 1,000 feet from the antenna of Station B, the desired signal was clearly readable to within 600 feet or less, and unintelligible at 300 feet.

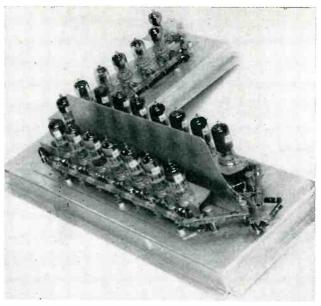
In the case of intermodulation interference resulting from the simultaneous transmissions of Stations B and C, although it was present in the absence of a signal from A, the latter completely captured the set up to a distance of about 800 feet from the antenna of Station B. Then there was a transition period where there was an intermingling of signals, and at approximately 600 feet, the intermodulation produced captured the receiver. The distance to Station C was approximately $1\frac{1}{2}$ miles in all cases.

These field tests proved that the design of the equipment (designated type MC203) had accomplished very satisfactorily its objectives of practical adjacent-channel operation with the maximum simplicity of circuits and lowest cost consistent with reliability and high-quality performance.

The authors wish to acknowledge their indebtedness to G. A. Kious and many others of their associates in the Radio Communication Engineering Section in the development, design and testing of the equipment.



A 150-mc distributed amplifier with a gain of 5.000 $\,$



An 85-mc distributed amplifier with a gain of 180

Millimicrosecond

Pulses as short as 0.5 millimicrosecond can be viewed or recorded through use of high-gain distributed amplifiers and high-vacuum linear-sweep generators. Distributed amplifiers show promise for utilization to 500 mc or higher

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OBSERVATION and measurement of pulses of the order of 10^{-6} to 10^{-10} sec have assumed increasing importance. This paper will discuss some recent oscillographic developments in circuits for measurement of these time intervals.

Development of high-speed sweep circuits^{1, 2}, refinements in reflection systems^{3, 4, 5, 6}, and waveform analysis using signal delay systems⁴ have been described in the literature. This paper is mainly concerned with the development of distributed amplifiers.

This article is based on a paper presented at the 1950 National Electronics Conference. The Conference paper will appear in *Proc. NEC*.

The development of amplifiers capable of satisfactory reproduction of signals containing signal components in the order hundreds of megacycles has been hampered by the limitations imposed by practical values of transconductance, interelectrode capacitances and input loading of vacuum tubes. The suggestion of increasing effective transconductance without increasing shunt capacitance by incorporating a number of tubes into a transmission line network was made many years ago, but only recently 6,9,10,11 has it been utilized to some degree. The principle of distributed amplification appears to

be the most promising for use with presently available tubes and components, and a number of practical designs have been evolved.

Distributed Amplifiers

Figure 1 shows the basic circuit of a distributed amplifier stage. Two artificial transmission lines are connected separately to plates and grids of the amplifier tubes, which become part of the lines. A signal voltage e_{ρ} applied to the input terminal is transmitted along the grid line to the grid of V_1 , producing a corresponding change in plate current $g_m e_{\rho}$, which is transmitted along the plate line. When

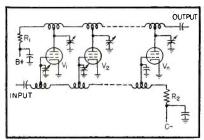


FIG. 1—Basic circuit diagram of a distributed amplifier

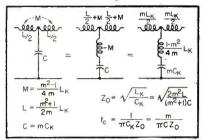


FIG. 2—Typical m-derived section and its equivalents

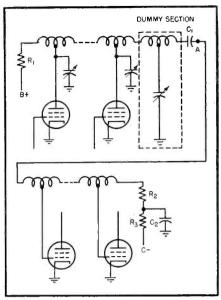


FIG. 3—Method of coupling two stages of a distributed amplifier

Oscillography

the applied signal reaches the grid of V_2 , a similar change of plate current will be produced in that tube.

If the propagation constants of these two lines are equal, the signal components of plate currents produced by various tubes will arrive at the output terminal at the same time and will add. The usual undesirable effects resulting from paralleling of tube capacitances are thereby eliminated.

When resistor R_1 is made equal to the characteristic impedance of the plate line, signal currents transmitted to the left will be completely absorbed. Similarly, the applied signal e_g will also be absorbed after the grid of V_n when R_2 is made equal to the characteristic impedance of the grid line.

Practical Considerations

For simplicity, both the grid and plate lines can be made of constant-k low-pass T-sections. However, m-derived sections have at least two advantages compared with constant-k sections. First, time delay can be made substantially constant over the entire pass band with less resulting phase dis-

tortion. Second, greater shunting capacitance can be used when m is greater than unity. This would permit either greater bandwidth or higher line impedance and correspondingly higher stage gain.

The m-derived network with substantially constant time delay up to two-thirds of the cutoff frequency is known to be that with m=1.27. A typical m-derived section is shown in Fig. 2 together with its equivalents. The negative inductance -M in series with the shunt capacitance C results from the aiding magnetic coupling between the two halves of L.

With L and C defined as in Fig. 2, the characteristic impedance becomes

$$Z_o = \sqrt{\frac{L_k}{C_k}} = \sqrt{\frac{2_m^2 L}{(m^2 + 1) C}}$$
 (1)

and the cutoff frequency is

$$f_c = \frac{1}{\pi C_k Z_o} = \frac{m}{\pi C Z_o} \tag{2}$$

The above expression indicates that the cutoff frequency of an m-derived section is m times that of a constant-k section.

Bridged-T networks offer the ad-

vantage of more constant characteristic impedance, but this is offset by the need for twice the number of critical circuit elements.

The signals traveling along both the grid and plate lines are subject to losses that will cause considerable attenuation at high frequencies because of the presence of impedance in the cathode circuit and the transit-time effect.

In a distributed amplifier stage the signal at the grid of the first tube is attenuated by these losses once, at the second tube grid twice, and so forth. An analysis assumes that the signal is attenuated by a db in the first section, by 2α db in the second, and so on by na db in the nth section. The effective sum of all signal components of plate currents is then reduced from $ng_m e_g$ on the output of an n-tube distributed amplifier stage to $g_{\scriptscriptstyle m} e_{\scriptscriptstyle g}$ $(a + a^2 + \ldots + a^n)$, where a =antilog $\alpha/20$. The value of this sum is $g_m e_g a(1-a^n)/(1-a)$ and the resulting loss in db is 20 log $a(1-a^n)/n(1-a)$. For a typical case of n = 7 and $\alpha = 0.6$ db, the resulting loss is 2.5 db.

Plate Line Losses

Similar consideration may apply to losses in the plate line due to low plate impedance of some tubes, such as the type 6AN5, whose plate impedance is likely to be below 10,000 ohms. The losses so caused, are perhaps 0.5 db per stage.

Because of these losses there is obviously a limiting number of tubes per stage beyond which the amplification of the stage will decrease instead of increase. A mathematical analysis assuming 0.6-db losses per section in both grid and plate lines shows the gain increases very slowly after n=8, becomes maximum when n=12 and decreases thereafter. The effect of coil losses due to skin effect and other factors is small compared with tube losses, about 1 db for 40 sections of 200-ohm networks.

Other factors such as grid and plate-lead inductances result in lowering the amplifier cutoff frequency and in altering the characteristic impedance of the transmission lines. Fortunately, it is possible to minimize such undesirable effects to a small percentage

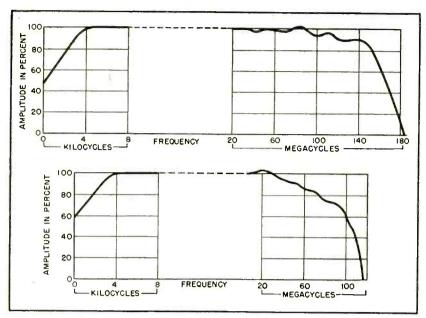


FIG. 4—Amplitude-frequency characteristics of distributed amplifier A (upper curve) and distributed amplifier B (lower curve)

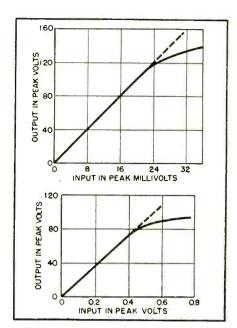


FIG. 5—Amplitude-amplitude characteristics of amplifier A (upper curve) and B (lower)

in a well-designed amplifier.

It is important to point out that uniform amplification with respect to frequency variation cannot be obtained in a stage of distributed amplification when the receiving end of the plate line is open. The reasons are: (1) The input impedance of a transmission line is no longer constant if it is not terminated by its own characteristic impedance, thus the line impedance looking at the plates of the amplifier tubes changes as the signal frequency varies; (2) The phase relations between the reflected wave and the signal components of various plate currents change as the signal frequency varies; therefore, they may add at one frequency and subtract at another.

Since the load of an oscillograph deflection amplifier is the deflecting plate of a cathode-ray tube, which is essentially a capacitive load of not less than 2 $\mu\mu$ f, it cannot be connected directly across the terminating resistor of the output line without introducing reflections.

To avoid this undesirable effect, an additional section is used, and the deflecting plates of the cathoderay tube are connected as part of the shunt capacitance. This capacitance is adjusted so that the total shunting capacitance is equal to that of other sections of the same line.

For amplification of pulses, the transient response of the amplifier is important, particularly the rise time and the percentage of overshoot. In order to minimize the overshoot of a multistage amplifier, it is necessary that the phase distortion be kept very low and that the frequency response drop slowly and steadily, following the Gaussian error curve. If f_1 is the frequency at the 70-percent point of the frequency response curve, the 37-percent point should be located approximately at $1.5 f_1$. The shape of the response below 30 percent is not critical, but it also should drop slowly and steadily.

Multistage Operation

The rise time of an amplifier stage is approximately equal to the ratio of 0.35/W, where W is the 3-db bandwidth of the amplifier. For nstages, the over-all rise time is approximately equal to the square root of the sum of the squares of those of the individual stages when the percentage of overshoot is negligible. It may be noted here that the effect of an oscillograph amplifier having rise time T_a on a signal pulse having rise time T_{p} can be estimated from the following expression:

$$T_{\text{resultant}} = \sqrt{T_{a^2} + T_{p^2}}$$
 (3)

Thus a signal pulse with a rise

time of 0.01 μ sec will appear to be slowed down by 11 percent in an oscillograph amplifier with $T_a=0.005~\mu$ sec to a resultant rise time of 0.011 μ sec.

Since there must always be certain physical distances existing between stages, the inductance of the leads employed for connecting stages affects the response of the amplifier. To overcome this difficulty, additional dummy sections are used. These sections serve as a continuation of the line, with tube capacitances replaced by the increased trimmer capacitances.

Distributed amplifier stages may be coupled directly in order that response may be extended down to d-c signals. For such operation. suitable means must be provided to overcome the difficulty of the presence of quiescent plate potential.12 When d-c response is not required. the arrangement shown in Fig. 3 has been found suitable to allow signals of very low frequencies to pass with the use of standard components and without affecting the response at high frequencies. physical size of the coupling capacitors should be kept as small as possible in order to avoid difficulties caused by both distributed capacitance and lead inductance of the coupling capacitor.

In Fig. 3, R_1 and R_2 are made equal to the characteristic imped-

ance of the line, C_1 is made equal to C_2 , and R_3 is made as large as possible for the type of amplifier tubes used. Under this condition, the impedance to the right of point A essentially equals that to the left except for the existence of a high resistance R_{s} . Theoretically, the frequency response will be uniform as long as the resistance of R_s is very large compared with the reactance of C_2 . Experimental results show that frequency response can be kept flat to as low as 200 cps even when the line impedance is 200 ohms and the coupling capacitance is only $0.01\mu f$.

Impedance transformation can be secured through the use of a stage of distributed amplification. For such operations the characteristic impedances of grid and plate lines are made respectively equal to the input and output impedances. When the stage consists of more than one tube, the propagation constants of both grid and plate lines should be made identical.

The amplification of a multistage distributed amplifier may be controlled by changing the number of stages or by inserting attenuators. The former may offer the advantage of better frequency response and shorter rise time, while the latter has been proved to be more practical in actual cases, since it minimizes the number of plug-in connections or switches required. While stage gain can also be controlled by variation of grid bias, this method is undesirable because it seriously affects the amplitude-amplitude characteristic, producing nonlinearity.

Results

Two multistage distributed amplifiers are shown in the photographs. Amplifier A consists of six stages, four single-ended stages, a phase inverter, and a push-pull output stage. Amplifier B consists of a single-tube stage used as impedance transformation for coupling to a 50-ohm coaxial cable, a single-ended stage, a phase inverter, and a push-pull output stage.

The individual stages of amplifier A were designed to have their theoretical cutoff frequency at 250 mc and their time delay approximately

 1.26×10^{-9} second, while those of amplifier B are at 125 mc and 2.52×10^{-9} second, respectively. The amplitude-frequency characteristics of these two amplifiers are shown in Fig. 4, and curves showing amplitude-amplitude characteristics of both amplifiers are shown in Fig. 5.

The rise time of amplifier A was measured to be approximately 0.004 μ sec and of amplifier B, 0.005 μ sec by the use of a special pulse generator capable of producing square pulses of 0.012- μ sec duration. The upper curve of Fig. 6

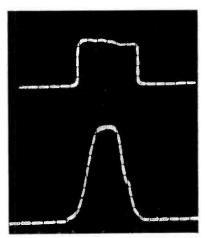


FIG. 6—Waveform of 0.012 µsec test pulse applied directly to vertical deflecting plates of cathode-ray tube (upper waveform) and same pulse after transmission through a 50-db attenuator and amplifier A (lower waveform)

shows the shape of the applied pulse as it appeared on the screen of a cathode-ray tube. The lower curve shows the shape of the same pulse after transmission through a 50-db attenuator and distributed amplifier A. Amplifier A has a gain of approximately 5,000 and is capable of delivering a balanced undistorted output voltage of 130 volts peak to peak. Amplifier B has a gain of approximately 180 and its maximum undistorted output voltage is 85 volts peak to peak, bal-

Table I—Technical Data for Distributed Amplifier A

Stages	Tubes	Function	Grid Line Ohms	Plate Line Ohms	Plate Current in Ma	Approx. Gain	Output Volts Peak- to-Peak
lst	8 6AK5's	Preamplifier	200	200	60	4	0.1
2nd	8 6AK5's	Preamplifier	200	200	60	4	0.4
3rd	8 6AK5's	Impedance					,
		Transformation	200	125	60	2.5	1
4th	8 6AN5's	Driver	125	125	270	6.5	6.5
5th	2 6AN5's	Phase Invertor	125	125	65	1	6.5
6th	24 6AN5's	Push-Pull Output	125	200	800	20	130
Total	58 Tubes		200	200	1.285 amp	5,000	130

Table II—Technical Data for Distributed Amplifier B

Stages	Tubes	Function	Grid Line Ohms	Plate Line Ohms	Plate Current in Ma	Approx. Gain	Output Volts Peak- to-Peak
lst	1 6AK5	Impedance					
		Transformation	50	400	9	2	2
2nd	5 6AK5's	Preamplifier	400	300	45	3.8	6
3rd	1 6CB6	Phase Inverter	300	300	15	1	2.5
4th	14 6CB6's	Push-PullOutput	300	520	210	23	85
Total	21 Tubes		50	520	279	175	85

High-Speed Sweep Circuits

anced about ground. A tabulation of technical data for both amplifiers

is given in Tables I and II.

Since the deflection of the cathode-ray beam is accomplished by a variation of electric field potential

between the horizontal deflecting plates, it is necessary to have this field potential changed at a rate dv/dt such that the spot will move at the required speed. For instance, a type 5XP cathode-ray tube operating with a second-anode potential of 4 kv and an intensifier potential of 25 kv requires approximately 300 volts change in electric field potential for one inch of horizontal deflection. The capacitance between these two plates, including wiring capacitance, is approximately 15 $\mu\mu f$. The current required for a sweep speed of 400 inches per μ sec becomes I = Cdv/dt = 2 amperes. This result indicates that a tube with high current capacity is required in a highspeed sweep circuit.

The requirement of high current capacity suggests immediately the use of a thyratron but time jitter is then produced on the screen of the cathode-ray tube. Furthermore, it is difficult to operate gas-tube sweeps at repetition rates above 10,000 or even 5,000 per second because of the relatively long time required for deionization.

If a high-vacuum tube is considered for use as a switch tube it will be required to conduct between successive sweeps in a conventional circuit. For high sweep speeds it becomes difficult to find a vacuum tube that can pass 2 amperes of plate current and that has interelectrode capacitances and lead inductances small enough for the highest sweep speeds desired.

A vacuum-tube sweep circuit in which the switch tube is normally

cut off and conducts only during the time of sweep is shown in Fig. 7. In this circuit the switch tube $V_{\scriptscriptstyle 5}$ is a vacuum tetrode type 4X150A. The lead inductances and interelectrode capacitances of this tube are quite small, yet its peak current capacity is over 2 amperes. If the repetition rate is kept under 100,000 per second, forced-air cooling of the tube is unnecessary.

Use of Tetrode

The reasons for using a tetrode instead of a triode as a switching tube are: (1) The input capacitance is substantially smaller because Miller effect is greatly minimized by the presence of a screen grid. This enables the sweep speed to become higher and simplifies somewhat the requirements of the trigger circuit. (2) Constant plate current can be obtained when the potential at the screen is made to rise or fall by the same amount as that at the cathode. Constant plate current is very important when linear sweep is required. In the circuit of Fig. 7 the potential at the screen of the switch tube is made to follow that of the cathode by the use of coupling capacitor C_4 .

Before the application of an input trigger pulse, all tubes are cut off. This results in low plate power consumption and permits the use of circuit components of low power rating.

When a trigger pulse is introduced to the grid of V_1 , plate current flows and a positive pulse is introduced at the grid of V_2 through magnetic coupling in trans-

former T_1 . Tube V_2 operates as a blocking oscillator. The time constant of R_1C_1 determines the width of the pulse produced by this blocking oscillator. The pulse is taken from the third winding of transformer T_1 to the grid of the cathode-ray tube for beam gating and to the input terminal of the delay cable for triggering the second blocking oscillator V_3 and V_4 . Use of the delay cable was found necessary in order to have the beam on at the instant when the sweep voltage starts. The time delay introduced by the cable of Fig. 7 is about $0.02~\mu {
m sec.}$ Components C_z and R_z are employed as a differentiating circuit.

The time constant of C_3 R_5 determines the width of the rectangular pulse on the grid of V_5 . Normally, tetrode V_5 is cut off, the voltage across capacitor C_5 is equal to about 2,000 volts and the voltage across C_5 is zero. During the pulse, the grid-to-cathode potential of V_5 is driven to zero or even slightly positive. Capacitor C_5 is discharged by the plate current of V_5 and capacitor C_5 is charged by the plate and screen currents of the same tube

Since the screen-to-cathode voltage of V_5 is kept constant by the use of coupling capacitor C_4 , both plate and screen currents are expected to be essentially constant during the pulse, thus permitting substantially linear sweep voltages to be developed across C_5 and C_6 .

The value of C_5 is slightly larger than that of C_6 because the current flowing to C_5 is larger than that

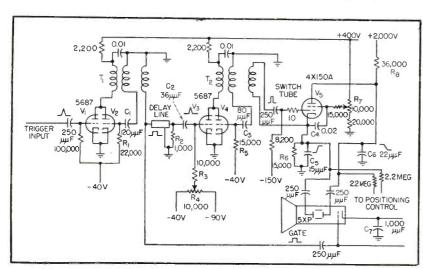


FIG. 7—Schematic diagram of high-speed sweep circuit

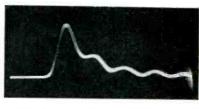




FIG. 8—Upper curve shows a 5 \times 10⁻¹⁰ second pulse displayed on a sweep generated by the circuit of Fig. 7. Lower curve is the same pulse and its reflection produced by a 30-inch open-circuited stub of RG-58/U cable

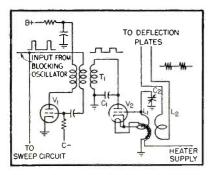


FIG. 9-Schematic diagram of high-frequency timing signal generator

flowing out from C_6 . After the pulse, tube V_5 is cut off, capacitor C_8 discharges through R_6 and capacitor C_6 charges through R_8 . The sweep speed can be changed by varying capacitors $C_{\scriptscriptstyle 6}$ and $C_{\scriptscriptstyle 6}$.

Experimental Results

Experimental results show that the circuit can generate a balanced sweep voltage of about 1,500 volts peak to peak. The maximum sweep speed obtainable is over 800 inches per μ sec when the sweep voltage is applied to the horizontal plates of a type 5XP cathode-ray tube with its second anode at 2,000 volts and third anode at 14 kilovolts. The speed is reduced by one-half when the second anode voltage is increased to 4,000 volts and the third anode is increased to 29 kilovolts. The sweep speed may be increased to 1,200 inches per microsecond when the sweep voltage is applied to the vertical deflecting plates of the cathode-ray tube.

The amount of time jitter is not more than the width of a normal trace on the cathode-ray tube and is estimated to be less than 0.0001 μsec. Departure from linearity was found to be less than 5 percent. The starting time from the instant of input trigger to the beginning of the sweep voltage is estimated to be $0.045~\mu sec.$ The maximum repetition rate is over 100,000 per second. The total consumption of plate and screen power is under 10 watts average.

Results of employing this sweep circuit to view a pulse signal of 0.5 millimicrosecond duration are shown in the curves Fig. 8. This pulse was generated by a circuit developed by the Purdue Research Foundation for the United States Air Force and the Signal Corps. The oscillations at the trailing edge of the upper pulse curve of Fig. 8 were produced by the pulse generator itself, since it employs the principle of reflection to generate narrow pulses. The lower curve of Fig. 8 shows the results of connecting a 30-inch open-circuited stub of RG-58/U coaxial cable at the signal deflecting plates. This arrangement will allow an accurate calibration of the time axis. Since RG-58/U coaxial cable has a characteristic impedance of 53 ohms and a capacitance per foot of 28 $\mu\mu$ f, the time delay will be 1.5 millimicroseconds per foot. Thus the time between the two major peaks of the lower curve of Fig. 8 is 7.5×10^{-6} second. The waveform of the pulse as shown may differ from the true waveform because of transit time effects in the cathode-ray tube and the frequency characteristic of the coaxial cable.

In order to calibrate the speed of the sweep, either a sine-wave os-

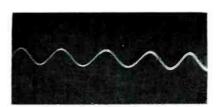


FIG. 10—A keyed 400-mc timing signal with a 2,500 prf, generated by the circuit of Fig. 9

cillator that oscillates at some known frequency or a pulse generator that produces sharp pulses at some known repetition rate can be employed to produce timing marks as vertical deflection. For sweep speeds of 400 inches per µsec, sharp pulses with rise times of less than 0.1 millimicrosecond would be required. Such pulses are difficult to generate and to apply to the deflecting plates or the control grid of the cathode-ray tube because of the leakage capacitances and inductances of all commercial circuit components involved. On the other hand, sine waves at frequencies of 1,000 mc or lower are comparatively less difficult to produce and to apply since the undesirable effects can generally be eliminated by the use of appropriate resonant circuits.

Synchronization of the timing signal to the sweep voltage is another important problem to be considered. For reasons described previously, the jitter should be less than 0.25 millimicrosecond in order to produce a satisfactory trace with a sweep speed of 400 inches per usec. It was found that a platepulsed oscillator can be arranged to satisfy these requirements.

Timing Signal Generator

The schematic diagram of such a circuit is shown in Fig. 9. In this circuit, the first blocking oscillator provides a trigger pulse to the sweep circuit and at the same time serves to trigger the second blocking oscillator V_1 . The output from the third winding of transformer T1 provides the plate current needed for oscillator tube V_z . The heater of V_2 is raised above a-c ground potential by the use of a choke wound on the lower half of coil L_1 . This is necessary because the cathode of V_2 is above a-c ground potential. Capacitor C2 is employed for frequency adjustment. The coupling between L_1 and L_2 should be adjusted experimentally until maximum deflection is obtained on the screen of the cathode-ray tube. Figure 10 shows the trace of a 400-mc sine wave generated by this circuit at a repetition rate of 2,500 per second on the screen of a type 5XP cathoderay tube.

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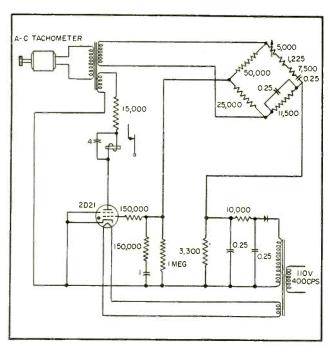


FIG. 1—Schematic diagram of electronic speed switch. The output frequency of an a-c tachometer is used as input to the frequency-sensitive network

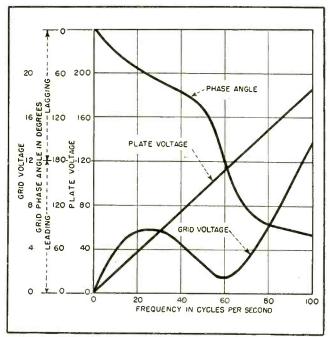


FIG. 2—Voltage and phase relationships for thyratron with Wien bridge arbitrarily balanced at 60 cps. Operating frequency of bridge is adjusted by the 5,000-ohm resistor

SPEED-CONTROLLED

Designed to operate from a standard a-c tachometer, an electronic switch makes or breaks contacts at a predetermined speed without materially increasing existing load. Switching on or off of auxiliary equipment can thus be done at the selected speed setting

SPEED CONTROL by the use of tachometers is less well known than speed measurement by the use of d-c and a-c tachometers. The simplicity of mechanical systems, such as a fly-ball governor whose output may be coupled directly to the speed-regulating mechanism, makes their use popular.

There are instances where an electrical system, particularly one using an a-c tachometer, offers distinct advantages over a mechanical system. An a-c signal whose frequency is accurately proportional to the speed of the rotating shaft can be generated without mechanical contact and brushes. In high-speed applications where gearing down to lower speeds for measuring purposes is not practical, the tachometer speed-control system may be

applied advantageously.

Remote indicators for speed are most likely to be electrical. If control signals at a certain speed are required, they can be derived from the same elements used for indication by use of a tachometer system.

Switch Operation

The electronic speed switch operates a relay at a preselected speed. The switch obtains its signal from an a-c tachometer whose output frequency is used for speed measurement. Figure 1 shows a schematic diagram of the speed switch circuit.

The tachometer output is applied through two different transformer windings to the plate of a 2D21 thyratron and to a Wien bridge. The output of the Wien bridge is connected to the grid of the thyratron. A negative d-c bias prevents the thyratron from firing unless voltages of proper phase are superimposed from the output of the bridge. The sensitive double-pole double-throw d-c relay in the plate circuit becomes energized when the thyratron conducts.

Figure 2 shows the plate and grid voltages of the thyratron and their phase relationship for a Wien bridge arbitrarily balanced at 60 cycles and a tachometer voltage which rises linearly with frequency.

If the frequency of the tachometer is equal to the balance frequency of the bridge, the grid voltage is 180 degrees out of phase with the plate voltage and the tube does not conduct. As the tachometer frequency increases, the grid volt-



Complete switch assembly. All electrical components are hermetically sealed in one container for greater economy in size and weight. Internal parts are supported by rubber bumpers

SWITCH

By H. W. KRETSCH and F. J. WALKER

Manning, Maxwell and Moore, Inc. Bridgeport, Conn.

age increases and leads the plate by an angle less than 180 degrees. Thus, somewhere at the beginning of each positive plate-voltage cycle the grid will become sufficiently positive to fire the thyratron for the balance of the half cycle and a current will flow sufficient to energize the relay.

Frequencies below the balance point will also cause a rise in the grid voltage, but in this case the grid voltage will lag the plate voltage by less than 180 degrees. The grid will be negative at the start of each positive plate-voltage cycle but will go positive at some point near the end of the half cycle. The average current created in this manner can be kept well below the value required to energize the relay.

The average current value might

become critical for very low frequencies were it not for the fact that the tachometer voltage also decreases with frequency. The relationship between grid and plate voltage at any one point can be obtained by the choice of the transformer winding feeding the bridge and plate circuit.

Even though a phase-controlled thyratron circuit has been chosen to make the operation dependent primarily on frequency, variations of the tachometer output voltage will influence the accuracy of the switch to some degree. A decrease in plate voltage, for instance, will require a larger grid voltage to fire the thyratron, as can be determined from the characteristics. To produce this larger grid voltage, a higher frequency is necessary which

in turn results in a change of the set point. For the values used in Fig. 2, a 10-percent change in tachometer voltage causes a shift of 1.2 percent in the operating point.

Line-Voltage and Temperature Effects

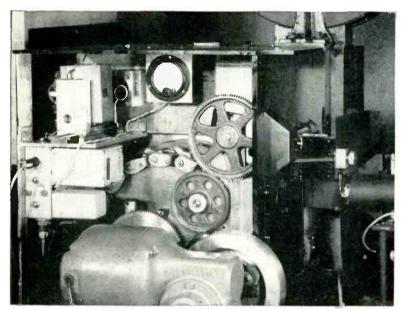
Line-voltage variations do not affect the operation of the switch because filament and bias voltage are both derived from the same a-c source. An increase in filament voltage decreases the operating point, opposite to the effect of an increase in bias voltage. With proper choice of the bias voltage the switch can be made independent of normal line-voltage variations.

To eliminate the influence of temperature variations, the 1,225-ohm bridge resistor is made of copper. Its value is determined so that it compensates for variations of the other bridge components and the thyratron to make the switch practically independent of ambient temperature variations.

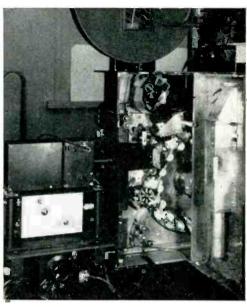
All electric components, including the relays, are hermetically sealed into one container to protect them from altitude, temperature and humidity variations, which are quite severe in aircraft applications for which this particular model was designed. Sealing all parts into one container, rather than providing individual hermetically sealed components, gives greater economy in size and weight, which is of utmost importance for aircraft installations. The chassis is connected to the container by means of rubber bumpers which furnish the necessary shock and vibration protection. One AN type connector has prongs for the dpdt relay as well as for the tachometer and power supply. Also accessible from the outside is a rheostat permitting a change in the speed setting,

Connected between the tachometer and speed indicator of an aircraft engine, this switch will actuate its relay at a predetermined setting without interfering with the indicating system. The power drawn from the tachometer to actuate the switch is not enough to disturb the indication. It can be used to operate auxiliary equipment at the selected speed setting without mechanical connection to the engine.

Continuous FILM



Motor drive, cam and rollers (left of center) of new projector. High precision is required on cam surface only over active angle of about 32 $^\circ$



Film transport, mirrors, gate (below center) and compensating mirror (above center)

CINCE the advent of the multiplier phototube in 1938, it has been recognized that the flyingspot method of televising transparencies, using a photomultiplier as the transducer, produces a higher quality signal than do most other pickup devices. One result of this situation is the fact that film transmissions from the BBC stations in England, using the flying-spot technique and a video bandwidth of 2.75 mc, are considered by most observers to be superior to film transmissions in the United States, using the iconoscope camera and a band of 4 mc. This marked difference in quality appears, despite the wider bandwidth available to American broadcasters, because of the limitations on contrast and background inherent in the storage action of the iconoscope, American technicians have sought for the past decade for a suitable nonstorage method of televising 24-frame-persecond film on the American scanning standard of 30 frames per second. In Great Britain, the difficulty is circumvented by running the film at a speed about 4 percent above normal, at 25 frames per second, which coincides with the European scanning standard.

One such group of technicians is the television research group of the Bell Telephone Laboratories, headed by Axel G. Jensen, which has developed a number of television film scanners to serve as sources of high-quality television signals for laboratory purposes. One such projector, demonstrated in recent years, used a special form of the Farnsworth image dissector (a nonstorage device) in conjunc-

tion with specially printed motion picture film. But since special film was needed, the range of subject matter available for testing purposes was restricted. To overcome this limitation, the group decided to design a continuous film projector, similar in some respects to the German Mechau machine now used by the BBC for some of its film transmissions, but containing a number of refinements which would

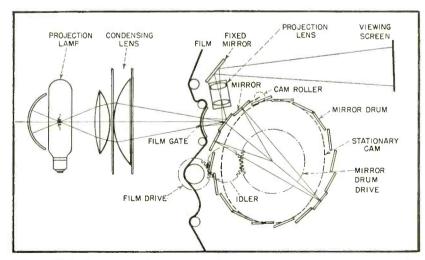


FIG. 1—Elements of projector when used for conventional projection on viewing screen. Roller attached to each mirror runs on stationary cam, turns mirror through angle to compensate for continuous motion of film past gate. Projected image contains vertical jitter and flicker

SCANNER for TV

Designed as high-fidelity television signal source by research staff of Bell Telephone Laboratories, mirror-drum projector uses two electronic servo controls to compensate for jitter and flicker, permits flying spot scanning of 24-frame film at 60 fields

adapt it to the American scanning standard and make its operation and maintenance relatively simple.

This new scanner, which is still under development, was described and demonstrated informally by Mr. Jensen and his associates before the recent convention of the SMPTE, whose members visited the Laboratories at Murray Hill on May 1. This account is derived from the information released at that meeting.

Nonintermittent Principle

Figure 1 shows the essential mechanical and optical elements of the new scanner, as it would be used for conventional projection of film on a viewing screen (rather than with a photomultiplier tube as described below). A projection lamp and condenser lens system pass light through a film gate of curved shape, over which the film passes. The film moves at constant speed, not intermittently as in conventional pulldown projectors. The image projected on the screen

would therefore be moving were it not for the fact that the optical path contains a series of mirrors in motion, their motion being such as to counteract the motion of the image.

The mirrors, eighteen in number, are mounted on the periphery of a drum or crown which rotates at constant speed, geared directly to the film drive at such a ratio that there is a crown mirror associated with each frame of the moving film. In other words, for 24 frame-persecond film, the crown rotates at 80 rpm.

Each crown mirror is free to rotate about an axis in the plane of its reflecting surface and at right angles to the plane of the diagram. Each mirror is caused to move about this axis by a lever attached to a roller, which rolls on the surface of a stationary cam. The shape of the cam is such that the motion of the mirror precisely counteracts the motion of the film, and a stationary image is thereby reflected through the projection lens to a fixed mirror and thence the viewing

screen as shown in Fig. 1.

Each mirror is active during about 32 degrees of motion. Since this is 60 percent more than the angular spacing of the crown mirrors, there is a lap dissolve of one frame into the next on the screen. In this manner, the image projected is continuous, one mirror alone being active for a portion of the time, two adjacent mirrors being active during the transition from one mirror to the next. When two mirrors are active the amount of light transmitted through the system changes somewhat, and this causes flicker at the film frame repetition rate of 24 per second. This flicker will be removed by an electronic servomechanism as described below.

Vertical Jitter

Aside from the flicker introduced by the successive action of the mirrors, this projector, in common with all other non-intermittent machines, is susceptible to vertical jitter, a slight upward and downward motion of the projected image with

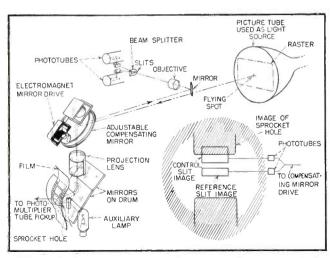


FIG. 2—Flying spot system and compensating servo which removes vertical jitter. Image of sprocket holes actuates mirror which rotates as required to compensate for any vertical motion of film frame as projected

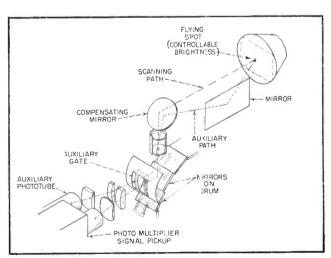


FIG. 3—Flicker-compensating servo. Light from flying spot is reflected from mirrors through auxiliary gate to phototube. Light passing through film changes through active cycle, but servo changes brightness of flying spot proportionately

respect to the frame edges. This effect arises mainly from the difficulty of maintaining absolute uniformity of film motion in the gate. This difficulty has been one of the outstanding objections to continuous projectors in the past, all of which have been found to require highly trained operators to maintain the mechanism in proper adjustment. In intermittent type projectors, this difficulty is avoided by the fact that the individual film frames are accurately aligned with the sprocket holes in the film, and these holes are accurately located, by the sprocket drive, with the respect to the film position in the gate during the dwell time of the intermittent motion. In this new projector, the vertical jitter is automatically corrected by means of an electronic servo which adjusts the position of the projected image by reference to an optical image of the sprocket holes. In addition, the servo decreases the accuracy required in the entire optical and mechanical assembly and alignment. It finally also tends to eliminate errors due to film shrinkage.

Jitter Compensation

Figure 2 shows the adaptation of the projector to the flying-spot method of pickup together with the elements of the jitter-correcting servomechanism. At the upper right is the picture tube (type 5ZP16) which serves as the source of light. The screen of this tube, which uses a very short persistence phosphor, is scanned in the usual manner (525 lines, 30 frames per second). The light from the scanning spot passes to the jitter compensating element, a front-surface plane mirror which reflects the light through a projection lens to the crown mirrors. These mirrors move past the film gate, and are turned by the cam mechanism through the proper angle to cause the spot of light to move downward with the film. The image of the raster is therefore stationary with respect to the film frame.

The light, having passed through the film, is collected and passed to a photomultiplier tube which develops a video signal amplitude proportional to the transmission factor of the film. As the scanning proceeds, the spot moves over the whole frame in the conventional series of alternate (interlaced) scanning lines, each line being stationary with respect to the film frame.

The jitter-compensating servo begins with an auxiliary projection lamp which illuminates, through a right-angle prism, the portion of the film containing the sprocket holes. The light is reflected by the surface surrounding sprocket holes. Consequently an image of the area of one of the sprocket holes, in motion, is reflected back to the crown mirrors, whose rotation arrests the motion in the manner described previously. Thus a stationary image of the sprocket holes is reflected back through the system, is picked off to one side as shown, and is imaged onto a pair of slits.

The light through one of the slits is controlled by the position of the sprocket hole edge, the light through the other slit is not affected by the edge position, and serves as a reference, as shown at the lower right of the diagram. The light from the two slits is fed through prisms to two phototubes. These in turn are connected to a differential amplifier whose output is proportional to the difference in light through the two slits.

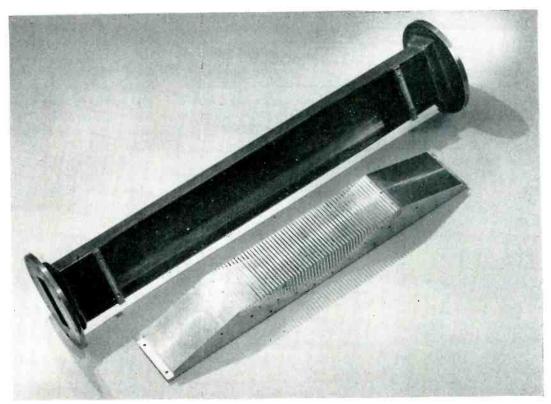
If a sprocket hole tends to get ahead of its correct position, due to jitter introduced by film or mirror drum, the light through the control slit increases and a control current of one polarity results, whereas if the sprocket hole falls behind, a current of the opposite polarity is produced. The control current, suitably amplified, is applied to a pair of electromagnets attached to the compensating mirror previously mentioned, at the left of the center of the diagram. The magnets cause the mirror to move in the proper direction and by the proper amount to shift the direction of the beams reflected by it so as to restore the edge of the sprocket hole to its proper position. This shift also corrects the direction of the beam from the flying spot to the film, and thereby maintains the raster image stationary with respect to the film

The degree of compensation is

such that the vertical jitter is reduced to less than 0.1 percent of the picture height, or less than half the width of a scanning line, which is well within the usual tolerance specified for scanning systems. To this performance, achieve servomechanism operates at high speed, and the compensating mirror moves through the required angle in a time of the order of a milli-Fortunately the angle second. through which the compensating mirror must rotate is of the order of a few minutes of arc, so the amount of driving power required is not excessive. In fact, the driving means consists of a pair of 5inch speaker voice coil units, and the peak power is of the order of one watt.

Flicker Compensation

Figure 3 shows a servomechanism proposed to compensate for the frame flicker previously mentioned. This feature was not a part of the apparatus demonstrated but no difficulty in incorporating it in the demonstration equipment was anticipated. The light from the flying spot is picked up by a mirror at the side which reflects the beam along an auxiliary path to the compensating mirror and projection lens and thence to the crown mirrors. The two beams, one on the regular scanning path, the other on the auxiliary path, fall on the mirror surfaces side by side as shown. Consequently the intensity of the two beams changes in the same manner through the active cycle of the crown mirrors. The auxiliary beam then passes through an open gate off to the side of the film gate and into a phototube which measures the light flux in the auxiliary beam. The phototube current, suitably transduced, is then applied as a control signal on the electron gun of the flying-spot cathode-ray tube, changing the intensity of the spot by an amount sufficient to compensate for the change in the light efficiency throughout the cycle. The servo thus acts to maintain the intensity of the light passing through the film at a constant level, and the video signal variations are representative only of variations of the transmission factor of the film being televised.—D.G.F.



Section of waveguide and corrugated surface which fits into it

Corrugated-Waveguide Band-Pass Filters

High-pass properties of a waveguide are combined with low-pass properties of a corrugated surface in a filter designed to give a rapid transition between pass and attenuation bands. Single corrugated element does work of several elements in conventional designs

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Rapid transition between pass and attenuation bands and a wide frequency range free of spurious responses are desirable properties of a band-pass filter. Such a filter can be realized readily by combining the high-pass properties of a waveguide with the low-pass property of a corrugated surface.

In a particular application a filter was designed which had a passband between 2,080 mc and 2,800 mc and 70-db attenuation, relative to the pass-band response, at 2,900 mc. There were no responses which were not attenuated by at least 60 db from 2,900 mc to above 10,000 mc.

Because of the high-pass characteristic of a waveguide, a band-pass filter can be formed readily by incorporating a low-pass structure into the waveguide. One such low-

pass structure is the corrugated surface for which approximate field solutions have been given in the literature¹⁻⁶.

The design considered in this paper for utilizing the corrugated surface structure is shown in Fig. 1, where one broad face of the waveguide is replaced by a tapered corrugated surface. Approximate solutions for the attenuation and phase characteristics of the filter

in Fig. 1 are obtained by combining design relations⁵ with a loaded-line equivalent circuit representation¹.

An idealized pass-band shape is illustrated in Fig. 2. The low-frequency cutoff f_c is provided by the normal cutoff of the waveguide, while the high-frequency cutoff $f\infty$, at which there is infinite attenuation, is provided by a resonance of the slots in region C of the corrugated surface.

The slots in region B of the corrugated surface serve a two-fold purpose. They provide additional resonant elements in the upper rejection band $(f>f\infty)$ to prevent repetitive pass bands. In addition they serve as a smooth taper which transforms the characteristic impedance of region C into the characteristic impedance of the waveguide through the taper in region A.

Filter Characteristics

The loaded transmission-line representation of one slot in the corrugated surface is shown in Fig. 3. The characteristic equation of the frequency spectrum relating the phase shift ϕ across one element of the loaded line (including both the slot and the transmission line of length ℓ' to the phase shift θ along ℓ' between two adjacent slots is

$$\cos \varphi = \cos \theta - (Z/2Z_0) \sin \theta \tan (p\theta) \ (1)$$
 where $\varphi = 2\pi l'/\lambda_s$
$$\theta = 2\pi l'/\lambda_g$$

$$p = (b - b')/2l'$$

$$Z = \text{characteristic impedance of the slot}$$

$$Z_0 = \text{characteristic impedance of the unloaded line}$$

$$\lambda_g = \text{guide wavelength in the unloaded line}$$

$$\lambda = \text{free-space wavelength}$$

$$\lambda_s = \text{guide wavelength in the loaded line}$$

$$\lambda_s = \text{guide wavelength in the loaded line}$$

 $\lambda_{\sigma} = \lambda/\sqrt{1 - (\lambda/2a)^2}$ (2) where a is the broad dimension of the waveguide.

Putting $\phi = \beta - j\alpha$ and neglecting resistive losses one obtains for the pass band $(\alpha = 0)$

$$\cos (\beta + 2 n \pi) = \cos \theta - (Z/2Z_0)$$

$$\sin \theta \tan (p\theta)$$
 (3)

where $n = 0, \pm 1, \pm 2$, etc and in the cutoff band $(\beta = 0 \text{ or } \pi)$

$$\pm \cosh \alpha = \cos \theta - (Z/2Z_0) \sin \theta \\ \tan (p\theta)$$
 (4)

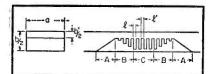


FIG. 1—Mechanical drawing of tapered corrugated surface which replaces one face of the waveguide

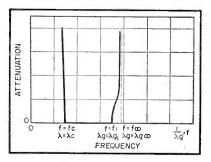


FIG. 2—Idealized pass-band shape. High-frequency attenuation at $f=\infty$ is provided by resonance of the slots in region C of corrugated surface (see Fig. 1)

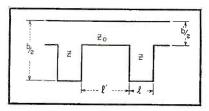


FIG. 3—Loaded transmission line representation of one slot in the corrugated surface

where $\alpha =$ attenuation in nepers per section of loaded line, $(\beta + 2n_{\pi}) =$ phaseshift per section of loaded line.

The values of $n \neq 0$ in Eq. 3 represent space harmonics of the fundamental (n = 0) component of the wave. The positive values of n represent transmitted waves, the negative values reflected waves. All of these components are necessary to fully describe the propagated wave.

To simplify the analysis, the space harmonic components are neglected. This is a valid assumption in the pass band, since the amplitudes of the space harmonics are greatly reduced in the pass band. The assumption becomes less valid as the frequency approaches that at which the slots become resonant.

Design relationships among the dimensions of a slot are

$$b'/2 < l' < (\lambda_{g 1})^2/10$$
 (5)

$$b'/b < 0.1 \tag{}$$

where $\lambda_{\mathfrak{p}1}$ is the guide wavelength in the unloaded section at the upper cutoff frequency, f_1 . It can be demonstrated that the passband resistive losses due to a finite conductivity in the conducting surfaces vary inversely with the quantity b'/b. This value was chosen to be as large as possible, namely 0.1. It may be seen from Eq. 4 that an infinite attenuation occurs when $p\theta = \pi/2$. The frequency at which this infinite attenuation occurs is $f\infty$. However,

$$\theta = 2\pi l'/\lambda_{g} = \frac{2\pi}{\lambda_{g}} \frac{(b-b')}{2p}$$

Hence at the infinite attenuation frequency $(\lambda_g = \lambda_g \infty)$

$$p \theta_{\infty} = \pi/2 = \frac{2 \pi}{\lambda_{g^{\infty}}} \frac{(b - b')}{2}$$

or

$$\frac{(b-b')}{2 \lambda_{\sigma^{\infty}}} = 1/4 \tag{7}$$

Eq. 7 means that at the infinite attenuation frequency, the slot depth, (b - b')/2, equals one quarter of a guide wavelength.

The values of l and l' are not critical and are generally chosen to keep the over-all length of the filter as short as possible. Typical values are $l'=2l=0.1\lambda_{cg}\infty$. The value of the ratio Z/Z_{\circ} is taken to be 2l/b', the ratio of the slot height to the line height.

Theoretical curves relating ϕ and α in db to Θ are shown in Fig. 4. The curves have been calculated for p=2 and $Z/Z_0=2$. At the point of infinite attenuation, $\Theta=45$ deg, while at the upper cutoff point $\Theta=36$ deg. Hence the ratio of the cutoff wavelength, λ_{gl} , to the infinite attenuation wavelength, $\lambda_{g}\infty$ is $\lambda_{gl}/\lambda_{g}\infty=45$ deg/36 deg = 1.25. The slot requires the cutoff wavelength to be decreased by 25 percent before its attenuation becomes infinite.

Referring to Fig. 4, representing the characteristics of the slots in region C of the filter, it can be seen that repetitive pass bands will occur. The first repetitive pass band occurs when $\lambda_{\sigma} = \lambda_{\sigma 1}/2$. To avoid this pass band as well as higher ones, it is necessary to include slots in region B of the filter which will resonate in the vicinity of the higher passbands. In general, this condition is fulfilled

merely by having a reasonable length of taper.

Design Procedure

The filter can be easily incorporated as an inserted section in a waveguide, with waveguide input and output. It can also be incorporated into a coaxial line through the use of waveguide to coaxial transformers. The transforming sections form the input and output of the filter.

The low-frequency cutoff determines the wide dimension of the waveguide, a, and is simply related to it by

$$\lambda_c = 2a$$

Generally, both the upper cutoff frequency f_1 and a minimum attenuation at a point between f_1 and $f \infty$ are specified. Once these are known, the slot depth and number of slots in region C can be determined. First λ_{g1} is calculated from the high-frequency cutoff f_1 by Eq. 2. Because $\lambda_g / \lambda_g \infty = 1.25$, $\lambda_g \infty$ may be readily found. The slot depth is then made equal to $\lambda_a \infty / 4$.

The number of slots required is the desired attenuation in db at some frequency between f_1 and $f \infty$ divided by the attenuation of one slot at that frequency as determined from Fig. 4.

The lengths of the tapered sections in regions A and B should be about equal and as long as possible for the best match. A theoretical curve relating the vswr introduced by a taper as a function of its length has been given by Frank'. From this curve and a consideration of the variation in guide wavelength in the pass band, a value giving small mismatch reflections over the entire pass band is found to be 1.5 λ_c or 3a.

Experimental Results

An experimental curve for the filter based on the previous design considerations is given in Fig. 5. The corrugated surface was designed to be inserted in standard 1.5 by 3-in. waveguide having a cutoff frequency of 2,080 mc. desired upper cutoff frequency was 2,800 mc ($\lambda_{g1} = 16.0$ cm.) This gives the infinite attenuation frequency as 3,200 mc ($\lambda_p \infty = 12.8$ cm). It was also desired to have

at least 70 db of attenuation at 2,900 mc ($\lambda_a = 15.0$ cm).

Because λ_m corresponds to $\Theta =$ 36 deg on the curve of Fig. 4, at $\lambda_a = 15.0$) cm, $\theta = 38.4$ deg. From Fig. 4, the attenuation in db per slot at $\theta = 38.4$ deg is 9 db. Hence eight slots in region C should be sufficient to give the desired attenuation. Actually ten slots were used because the attenuation of a slot will be slightly less than shown in Fig. 4 due to resistive losses in the slot. The pass-band loss, averaging approximately 2.5 db, is a combination of resistive loss in the slots and mismatch loss in the tapers.

Higher Order Modes

In the above analysis it was assumed that only the dominant mode $TE_{1,0}$ is propagated. For the usual waveguide dimensions (broad face twice as wide as the narrow face) higher-order modes may be propagated at frequencies greater than twice the normal cutoff frequency. If these higher-order modes are propagated, spurious responses may appear in the rejection band.

Because of the symmetry of the filter, only $TE_{i,in}$ $(n = 0,1,2,\ldots)$ and $TM_{1.2n}$ (n = 1,2,3,...) modes can be excited within the filter itself. The first of these modes, the

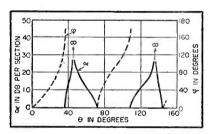


FIG. 4-Theoretical curves relating attenuation in db per section and phase shift across one slot for transmission line of length I' (see Fig. 1) to the phase shift along the distance between two

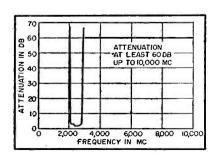


FIG. 5-Attenuation curve for actual filter constructed

 $TE_{1,2}$ and the $TM_{1,2}$ will not be propagated for frequencies less than four times the normal cutoff frequency. In general this is far enough into the rejection band to be of little consequence.

Should higher-order modes, such as the $TE_{2,0}$, $TE_{3,0}$ be set up in the input section leading to the filter, they will pass through the filter and produce narrow spurious pass bands for frequencies in the vicinity of twice the normal cutoff frequency, three times this frequency, and so on.

At the cutoff frequency for the higher modes, the guide wavelength for these modes is infinite and rapidly decreases as the frequency is increased. Until the wavelength decreases to a value such that the slots in region C become resonant, no attenuation is offered to the modes.

The higher modes can often be eliminated by careful design of the input circuit to eliminate asymmetrical structures tending to excite the higher-order modes. They may also be eliminated in the output section when of a special type, such as ridged waveguide output. In this case, the ridged section is designed so that it passes the dominant mode, but is cut off for the higher-order modes in the vicinity of the spurious response pass bands. Compensating sections of different widths, a, can be included in the filter proper so that they are below cutoff for the higher-order modes over the frequency range of the spurious responses8.

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REFERENCES

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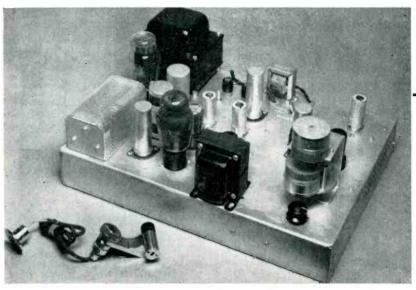
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BROADCAST TAPE



Complete equipment except for audio amplifier to drive capstan motor. Scanning unit cylinder at lower left is 34 inch in diameter and 2 inches long

THE MANY conveniences of magnetic tape as a medium for recording radio programs are attractive to producers and artists, as well as broadcast technicians. A complete show can now be recorded in a manner similar to the production of a Hollywood movie, with each selection recorded over and over until the director is entirely satisfied. The producer can then edit the recordings to produce precisely the type of show he desires. Fluffs may be easily removed and replaced. This practice provides a freedom of expression for the artist that is extremely convenient. During the editing process the show can be carefully timed so that it will be completed without sacrificing valuable commercial time.

To insure desirable continuity, the program must be timed within one second in a half hour. The dimensional stability of plastic-base tape is not sufficient to guarantee that the recorded time will be maintained to such close tolerances during reproduction of the program.

Conditions of temperature and humidity as well as tension affect the physical dimensions of plastic materials. These dimensional changes may increase with time. Therefore, programs that are shipped to other locations having different climatic conditions or those stored for long periods of time may suffer severely in this respect. In such cases, the time required for reproducing the program may vary considerably from the recording time.

Timing Controls

Several different methods of controlling the playing time of tape recordings have been suggested or tried. One utilizes transverse magnetization to record a reference signal on the tape during the recording of the program. This effect is accomplished by rotating the signal recording head 90 degrees so that the gap is placed parallel to the direction of tape travel. The signal recorded in this manner is reproduced during playback by a similarly oriented head and is used for synchronizing purposes.

A second system² makes use of a 14-kc carrier that is modulated with a timing signal. During reproduction, the carrier is filtered from the program signal and after demodulation is used for synchronization.

The system we have developed

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utilizes optical markings to furnish the timing signal. Since the signal markings are not magnetic, no unwanted signal is generated in the reproducing head by the additional information stored on the recording medium.

The signal is produced by equally spaced bars, printed on the reverse side of the tape, as they are scanned with an optical transducer. This transducer comprises an extremely simple optical system. A grid. placed directly over the printed side of the tape, consists of a mask with one or more holes to correspond with the markings on the tape. A light source is directed towards the grid and as the tape changes in color beneath the hole in the grid. more or less light is reflected back into a photocell.

Figure 1 shows a cut-away view of a scanner to illustrate the operation of such an optical system. Part of the light from source B goes through the aperture or grid C onto the tape A. The intensity of the light reflected back to the photocell D is partially governed by the color and gloss of the tape surface. The printed markings therefore produce a signal in the photocell whose frequency is directly proportional to the speed of the tape past the aperture.

The signal generated by the motion of the tape must be compared with some standard or reference frequency, which may be obtained from a regular power supply or from a stable clock-circuit power supply. Some large broadcasting studios have the latter to insure the accuracy of clocks used for program timing. Since available clock motors are designed to operate on 60 cps and the NAB standard tape speed is fifteen inches per second,

SPEED CONTROL

Magnetic-tape programs can be played back within less than a second of original half-hour recording time despite 2-percent stretch or shrink. Printed bars on back of tape are scanned photoelectrically to produce a comparison signal that is held at exactly 60 cps

the markings on the tape are placed at four bars per inch. If it is desirable to operate the tape at the secondary standard of 7.5 inches per second, the marking of four bars per inch may still be used to produce a 60-cps signal by means of a frequency-doubling circuit.

Comparing Speeds

Several methods of comparison between reference and photocell frequencies could have been used. One utilizes a simple beat note for driving the capstan motor. The signal from the photocell is mixed with a signal of twice the reference frequency and the beat note produced will then vary in frequency inversely with the frequency of the photocell signal. This system decreases the error but can never introduce sufficient compensation for complete correction.

A second method uses a twophase motor as an integrator. The signal from the reference frequency is used to furnish power for one winding of a two-phase capstan The signal furnished by the motion of the tape is amplified to excite the second winding of the motor. If the phase of the current through the first winding leads that of the current in the second winding, the torque of the motor is increased, and conversely, if the phase of the current through the second winding leads that of the current in the first winding, the torque of the motor is decreased. Such a system, however, is an instantaneous or phase-integrating device and does not furnish the time-integrating system desirable for broadcast timing.

Small, fast deviations may produce extremely high percentages of wow. This fact makes it imperative that any correction in speed be accomplished at a very slow rate.

A third method providing a slow

time-integrating device may be designed using two synchronous clock One operates from the clock power supply, using no more than 4 watts from this source. A second clock motor operates from the signal generated by the photocell, with rotation counter to that of the first motor. The shafts of the two clock motors are joined by a flexible coupling. As long as the two frequencies governing the speed of the two motors are identical nothing happens, but any difference in the two frequencies causes the frame of one of the motors to rotate with respect to the frame of the second one. This might be described as a simple mechanical differential, which needs no special gears or precision parts.

Capstan Speed Correction

The rotation of the clock frame may be utilized in a number of different ways to obtain correction of

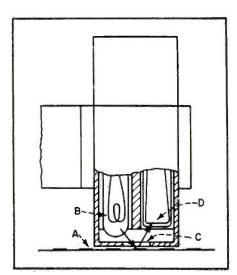


FIG. 1—Light from lamp B reaches photocell D by reflection from tape A

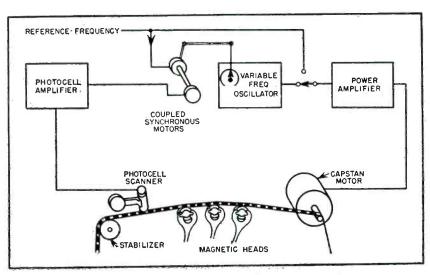


FIG. 2—Block diagram of the final system in which closed-circuit servo operates capstan motor so that signal from tape is exactly 60 cps

the capstan speed. One scheme is mechanically to shift a belt sideways on a tapered pulley. Such a method provides surprisingly accurate correction for timing purposes.

A second mechanical method may be arranged by driving a disk with a friction roller. Speed corrections may be introduced by shifting the roller radially with respect to the disk. The timing with such a system may be controlled very satisfactorily. Such a method is again handicapped with inherent mechanical difficulties that introduce wow in the program material. The precision necessary in the construction of such a device to overcome these difficulties would probably make the cost prohibitive.

Any type of mechanical system is handicapped because the mechanical power developed by standard clock motors is extremely limited.

Overdrive System

The dual clock-motor integrating system may be used satisfactorily by providing an electronic correcting system. One means of doing this is to have the speed of the capstan slightly more than required to drive the tape at the desired speed. The capstan may then be slowed down to the proper speed by an electronic brake. The power de-

livered to the electronic brake may be controlled by the rotation produced by the dual clock-motor differential.

There is one serious difficulty encountered in using any type of overdriven system. If, for any reason, it is desirable to record or reproduce tape without synchronization, the machine will be operated at some speed that is higher than normal, making the system incompatible with standard unmarked tape. This incompatibility can be eliminated by driving the capstan with a synchronous motor as shown in Fig. 2. The motor is driven with a power source of variable frequency when utilizing the synchronizing system, or with fixedfrequency power at other times.

The motion of the dual clockmotor differential is used to vary the frequency of a local oscillator having a mean frequency of approximately 60 cps. The signal from the oscillator is amplified with an inexpensive narrow-band power amplifier to drive the capstan motor. A relay is also incorporated so that the input signal to the power amplifier is derived from the local oscillator when the optically marked tape is used. If standard unmarked tape is used, the signal that drives the capstan motor is automatically derived from the clock circuit or other exact reference frequency.

The circuit diagram of the photocell amplifier is shown in Fig. 3. The exciter lamp in the scanning unit must be heated with either direct current or alternating current having a frequency that will not interfere with the operation of the clock motor driven by the signal from the photocell.

Exciter Voltage

Direct current for the exciter lamp might be obtained simply by inserting it in series with the amplifier power supply. However, if such a method is used, it is extremely difficult to filter the power supply sufficiently to eliminate low-frequency feedback effects or motor-boating. This difficulty was avoided by using a high-voltage, low-current lamp and connecting it in the bleeder circuit of the d-c power supply.

This particular lamp was designed to be operated with 55 v and 0.047 amp. Many such lamps are now available requiring currents as low as 0.032 amp.

The photocell selected is a lead sulphide type of small physical dimensions. The glass bulb of the particular one chosen measures only § inch long and ½ inch in diameter. Cadmium sulphide cells are available in even smaller sizes (approxi-

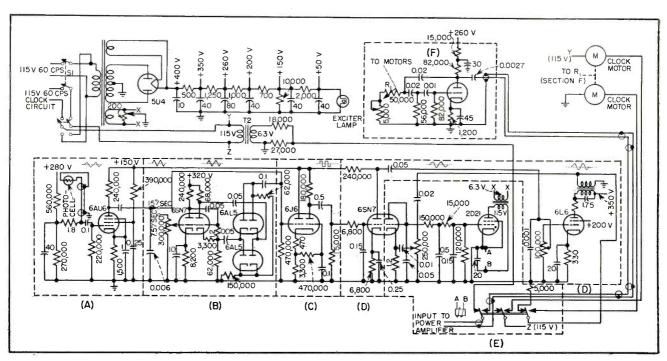


FIG. 3—Complete circuit diagram showing approximate waveforms and their relative amplitudes. Circuits in the dashed boxes are individually described in the text

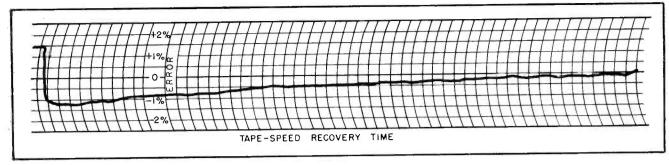


FIG. 4—Recovery time after drastic intentional error was introduced into the system

mately \(\frac{1}{2}\)-inch cube). The lead sulphide type was chosen also because of its sensitivity at the red end of the visible spectrum.

Since all the better quality plastic-base tapes are coated with ferric oxide, their color is a reddish brown. The markings can be printed on this color with black ink and produce good contrast when scanned with a red-sensitive photocell. If black tape is used with white markings, the contrast is also sufficient with the same photocell.

The lead sulphide photocell has an internal resistance that varies inversely with the light applied to its sensitive area. The circuit requires a voltage supplied through a load resistance or inductance in the same manner as the anode of a vacuum tube.

The output from the photocell is fed to a conventional preamplifier (Fig. 3A). If the tape is being operated at half speed or 7.5 inches per sec, the signal is diverted through a frequency-doubling circuit (Fig. 3B). This circuit consists of an amplifier, to regain the losses caused by the other stages; an isolation stage, to provide balanced input for the bridge; and a full-wave bridge-type rectifier, to provide the frequency doubling. The gain of the amplifier is so adjusted that the output signal has approximately the same level as that of the input but at twice the frequency.

The output from the frequency-doubling circuit or preamplifier, as the case may be, is passed through a limiter or clipper circuit (Fig. 3C). This circuit provides an output of constant amplitude even though the input signal may vary as much as five to one.

The output from the clipping cir-

cuit is then amplified sufficiently to operate one of the clock motors (Fig. 3D). Since the signal is changed to approximately a square wave in the clipping circuit, it is necessary to reshape the signal in the amplifier circuits. A nearly sinusoidal signal is provided by resistance-capacitance filters as well as by a capacitor tuning the inductance of the clock motor itself. Changes in the waveform of the signal are indicated on different sections of the diagram. Signals from the photocell and preamplifier are shown for a tape speed of 15 in. per second.

Control Relay

The relay that stops the clock motors and switches the signal for the capstan motor is electronically controlled (Fig. 3E). The coil is in series with the cathode circuit of a thyratron tube and the relay is normally energized. When the grid of the thyratron is biased negatively, the tube is cut off and the relay is denergized.

Bias for the thyratron is furnished by rectifying some of the signal generated by the photocell. The signal is furnished to the diode rectifier through a high-pass filter network from the output of the power amplifier stage. This filter network may be adjusted so the relay will become de-energized as the frequency of the photocell signal approaches 60 cps.

The variable-frequency oscillator (Fig. 3F), furnishing the signal for the capstan motor, uses an R-C circuit of the phase-shift type. Three R-C phase-shifting elements are used, with each unit shifting the phase approximately 60 degrees. One of the resistance elements is varied with the motion developed by the dual clock-motor

differential, and so drifts the frequency of the oscillator.

It may be noted that the impedances of the R-C circuits are kept as low as possible. This precaution was used to decrease the possibility of hum pickup in the grid circuit. Any oscillator has a tendency to lock-in on any signal appearing on the grid of the tube. If lock-in occurs, the usefulness of the entire system will be jeopardized.

No circuit is shown for the power amplifier needed to drive the capstan motor. A conventional audio power amplifier may be used or a much simpler narrow-band amplifier could be constructed.

Performance

Repeated timing tests have been made with this synchronizing system. A synchronized tape was recorded with timing markers exactly 15 minutes apart. Reproduction of this tape was again timed. The timer was operated from the same source of power as that furnishing the reference frequency in both cases. The greatest differential between recording and reproducing time in several tests was 0.3 second. This error might be reduced still further at the expense of speed variation or wow. A recovery rate was selected so that any speed changes would occur as slow drift and not be detected as wow.

The curve in Fig. 4 depicts the recovery rate of tape speed after an intentional error that had been introduced into the correcting device by manually rotating the clockmotor differential.

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Measuring Vector Relationships

Imaginary and real components of an unknown voltage in terms of a reference voltage are indicated directly over a frequency range of 8 cycles to 500 mc. Other applications include study of phase delay in amplifiers, wave filters and attenuators

Vector Relations of alternating voltages over a wide frequency range, from 8 cycles to 500 megacycles, can be measured by means of the instrument to be described. In most applications, the operating principle is based essentially on the vector theory of addition and subtraction. The operating principles underlying two frequent applications, measurement of the phase angle between two given voltages and measurement of complex components of an unknown voltage, will be discussed.

Phase-Angle Measurement

Figure 1 shows a block diagram of the instrument as it is used for the measurement of the phase angle between two given voltages, E_1 and E_2 . Variable attenuators R_1 and R_2 are used for adjusting the amplitudes of E_1 and E_2 . Phase inverters P_1 and P_2 are capable of producing a constant phase shift of 180 degrees over the entire range of operating frequency.

When terminals 4 and 8 are connected to terminals 1 and 6 respectively, output meter V will read the vector difference of E_1 and E_2 . If the absolute amplitudes of E_1 and E_2 are made equal by adjusting R_1 and R_2 , the reading E_0 on the output meter may be expressed as

$$E_o = E_2' - E_1'$$

= $|E_1'| (\cos \theta + j \sin \theta - 1)$ (1)
where Θ is the phase angle between

By Y. P. YU
Passaic, N. J.

 E_1 and E_2 . The absolute magnitude of E_a may be written as

$$\begin{aligned} |E_o| &= |E_1'| \sqrt{(\cos \theta - 1)^2 + \sin^2 \theta} \\ &= 2|E_1'| \sin \frac{\theta}{2} \end{aligned} \tag{2}$$

For simplicity, the magnitude of $|E_{\scriptscriptstyle \perp}'|$ is made equal to one volt dur-

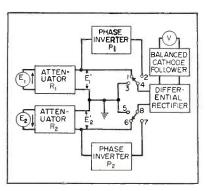


FIG. 1—Block diagram of the Vectorlyzer

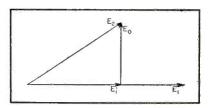


FIG. 2—Vector diagram used in connection with measurement of complex components of an unknown voltage

ing adjustment of attenuator R_1 , whereby the term $|E_1|$ may be omitted in Eq. 2, and

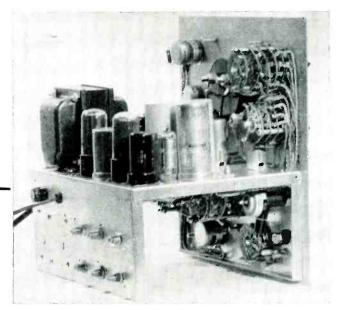
$$|E_o| = 2\sin\frac{\theta}{2}$$

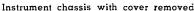
The output meter may be calibrated to indicate directly the phase angle in degrees between any two voltages. For instance, a meter with full-scale sensitivity of two volts will read 60 degrees at its center and 180 degrees at full scale.

The distribution of readings on a meter face is an important factor in the design of measuring instruments. A linear-scale meter which has a guarantee of one percent error at full scale would produce an error of 10 percent at one-tenth full scale. Sinusoidal distribution of meter readings is a desirable feature since the percentage accuracy of readings is nearly uniform over the entire scale.

Where the two voltages under consideration are nearly opposite in phase, the accuracy of reading can be greatly increased by introducing a phase shift of 180 degrees to either one. This is done with phase inverter P_1 or P_2 . A second scale starting with 180 degrees at its normal zero position is established.

Without altering the value of any circuit element the full-scale sensitivity may be increased to any value between 0 and 180 degrees or between 180 and 360 degrees by increasing the value of $|E'_1|$.







Front-panel view of the complete instrument

The principle underlying this feasibility can best be explained by referring to Eq. 2. As the phase angle θ of this equation decreases, the magnitude of the resultant voltage $|E_{\phi}|$ may be held constant if the amplitudes of the input voltages increase accordingly. This in turn keeps the indication on the output meter unchanged, since it is energized by the resultant voltage $|E_{\phi}|$.

If it is desired to increase the full-scale angular sensitivity of the instrument from 180 degrees to 60 degrees with the voltage sensitivity of the output meter remaining at 2 volts, R_1 and R_2 are readjusted so that the magnitudes of $|E_1|$ and $|E_2|$ will be two volts instead of one volt. After this readjustment the angular sensitivity of the output meter will be increased by three times.

Measuring Complex Voltages

The instrument is a simple and convenient device for direct indication of the imaginary and real components of an unknown voltage in terms of a reference voltage. The block diagram of Fig. 1 may also be used to explain the operating principle underlying this application.

In Fig. 1, E_1 and E_2 are assumed to be the reference and unknown voltages respectively. Attenuator R_2 is adjusted to give zero attenuation so that E_2 will be equal to E_2 .

Attenuator R_1 is adjusted until the deflection at the output meter reaches a minimum with terminals 4 and 8 connected respectively to terminals 1 and 6. To represent this condition, a vector diagram is drawn in Fig. 2 for voltages E_1 , E_2 , and E_o . Symbol E_o denotes the reading on output meter V, which is also the potential developed across terminals 1 and 6 when the meter is correctly calibrated.

Since these three vectors form a right triangle, the diagram indicates that E_{\circ} is equal to the imaginary component of E_{\circ} and that E_{1}' is equal to the real component of E_{2} . The output meter will read directly the imaginary component with terminals 4 and 8 connected to the positions shown in Fig. 1 and read directly the real component when terminal 8 is connected to terminal

If the phase angle between the unknown voltage and the reference voltage is greater than 90 degrees but less than 270 degrees, the above method of measuring complex components still is useful when E_1 is shifted 180 degrees with respect to E_1 by inserting phase inverter P_1 .

High-Impedance Attenuator

The attenuator circuit shown in Fig. 3 is designed to minimize phase-distortion difficulties encountered with conventional volume controls at high frequencies. The tri-

ode is arranged as a cathode follower to provide a high input impedance and to allow the resistance of the potentiometer at its cathode to become very low. The first feature is essential for many practical applications while the second helps to minimize the error caused by the stray capacitances of the potentiometer.

Experimental results show that the maximum error introduced by the circuit of Fig. 3 is less than 2 degrees at 2.5 mc with a 1,000-ohm carbon potentiometer for R_* . At low frequencies, the error angle is determined by the values of C_2 and R_s —the larger these two elements are, the smaller the error angle. Capacitor C_2 keeps the d-c potential constant at the output terminal during adjustment.

Figure 4 shows the schematic diagram of the differential rectifier and the balanced cathode-follower circuits appearing in Fig. 1. Diode V_1 functions as a differential rectifier. During the positive loop of the applied voltage, C_1 and C_2 charge through V_4 . During the negative loop of the applied voltage these two capacitors discharge through the series combination of R_1 , R_2 and the other part of the circuit.

The time constant during discharge is made very large compared with that during charge, and therefore the sum of the average poten-

tials across C_1 and C_2 is only slightly less than the peak value of the positive loop of the applied voltage. Resistors R_1 and R_2 are used to isolate both input terminals from the other part of the circuit. Both input terminals can be connected to points which are above a-c ground potential.

Triodes V_2 and V_3 function as a balanced cathode follower. The impedance between points C and Dlooking into the cathode follower is $R_3/(1-A)$, where A is the gain of the cathode follower. This impedance becomes very large compared with R₃ when A approaches unity. In practice, the value of $R_3/(1-A)$ can be conveniently made equal to 100 megohms or higher if R_3 is chosen to be 10 megohms. This high impedance makes it possible to operate the instrument at frequencies as low as a few cycles since the time constant $[C_1C_2/(C_1+C_2)]$ $[R_3/(1-A)]$ determines the lower limit of the frequency response.

Another feature of the circuits described is good stability. This is achieved by the use of balanced arrangement. Physically, R_5 , R_8 , V_1 - V_2 and V_3 - V_4 perform as four arms of a balanced bridge in which R_5 and R_8 correspond to the ratio arms, V_1 - V_2 to the unknown arm and V_3 - V_4 to the standard arm.

VHF Operation

With available circuit elements and technical knowledge it is extremely difficult to construct a variable attenuator which will produce negligible or constant phase shift from 10 to 500 mc while also having a high input impedance. It is equally difficult to construct a high-impedance network which will introduce a constant phase shift over this range.

Figure 5 shows the basic circuit of the instrument when it is used at high frequencies. In this circuit, rectification is performed by the probe which consists of a germanium crystal, two capacitors and two resistors. The resonant frequency of the probe can be made very high, over 700 mc, when care is taken during construction. Potentiometer R_4 is used for panel zero adjustment. Before operation the variable arms of potentiometers

 R_s and R_{θ} are set at their extreme right positions.

In order to explain the operation of this circuit, consider the case of measuring the phase angle between two given voltages. Assume that symbols E_1 and E_2 represent the two given voltages and the magnitude of E_1 is smaller than that of E_{2} . The procedure for handling various controls is as follows: (1) Connect input terminal 6 to ground and connect terminals 4 and 5 to terminals 1 and 2, respectively. Adjust potentiometer R3 until a halfscale deflection is obtained on the output meter. This step serves to adjust the sensitivity of the instrument so that a full-scale deflection will occur if the phase angle to be measured is 180 degrees. (2) Connect terminal 4 to ground and terminal 6 back to terminal 3, then adjust potentiometer R, until a halfscale deflection is again obtained on the output meter. This step serves to adjust the zero indication of the instrument so that the output meter will read zero when the two input voltages are in phase. (3) Connect terminal 4 back to terminal 1. Now enter the reading of the output meter into one of the curves of Fig. 6 and find the answer on the X axis. If the absolute magnitudes of both input voltages are equal, the phase angle can be read directly on the output meter and it is possible to omit both the second step and the use of the curves of Fig. 6.

To explain the operating principle, an expression for the voltage E_o developed across terminals 1 and 6, which is also the reading on the output meter, is written as follows:

$$E_o = E_2 - E_1 = |E_1|(n\cos\theta + j n\sin\theta - 1)$$
 (3)

where $n = |E_2|/|E_1|$ and Θ is the phase angle between E_1 and E_2 . The absolute amplitude of E_a is

$$|E_o| = |E_1| \sqrt{n^2 + 1 - 2n \cos \theta} \tag{4}$$

Equation 4 indicates that $|E_o|$ varies from (n-1) $|E_1|$ to (n+1) $|E_1|$ as the angle Θ varies from zero to 180 degrees. Thus the maximum variation is 2 $|E_1|$ which is independent of the ratio n. This relation is very important in using the instrument at high frequencies.

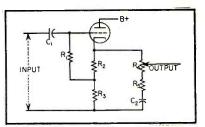


FIG. 3—Basic circuit diagram of the attenuator for audio and radio frequencies

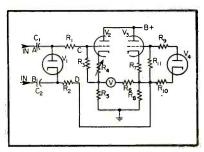


FIG. 4—Differential rectifier and balanced cathode follower

Assume that Θ_1 is the reading of the output meter and Θ is the true phase angle between E_1 and E_2 . Since the scale of the output meter is calibrated in degrees for the case when both input voltages have equal amplitude, we may write

$$2|E_1|\sin\frac{\theta_1}{2} = |E_1| \left[\sqrt{n^2 + 1 - 2n\cos\theta} - (n-1) \right]$$
 (5)

In Eq. 5, the term at the left is obtained by substituting n=1 into Eq. 3, while the term at the right is derived from Eq. 4 and the fact that a voltage with amplitude equal to (n-1) $|E_1|$ is balanced out from the output meter during the second step of adjustment. By rearrangment of terms, Eq. 5 becomes

$$\cos \theta = 2\left(1 - \sin\frac{\theta_1}{2}\right)\left(1 + \frac{1}{n}\sin\frac{\theta_1}{2}\right) - 1 \quad (6)$$

Graphical representation of Eq. 6 is shown in Fig. 6.

Complete Instrument

A simplified circuit diagram of the complete instrument is shown in Fig. 7. Essentially the instrument is a combination of the following elements: two variable attenuators, two phase inverters, a differential rectifier, a balanced cathode follower, a high-frequency probe,

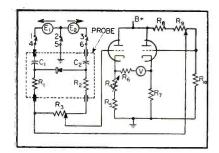


FIG. 5—Circuit diagram of the Vectorlyzer for operation at high frequencies

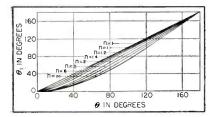


FIG. 6—Curves used in conjunction with the high-frequency probe

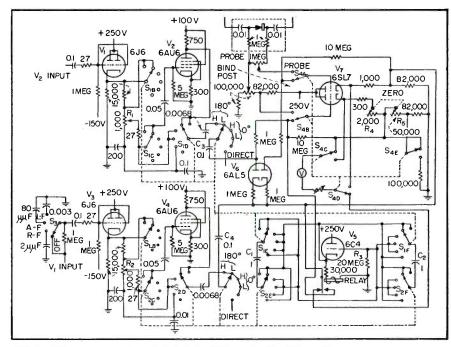


FIG. 7—Schematic diagram of complete instrument. Duo-triodes V_1 and V_3 are employed as cathode followers in the attenuator circuits

an automatic relay circuit and a regulated power supply. Duo-triodes V_1 and V_3 , having a transconductance of approximately 10,000 micromhos with both sections connected in parallel, are employed as cathode followers in the attenuator circuits, permitting the use of low-resistance potentiometers R_1 and R_2 for adjustment.

Two single-stage degenerative amplifiers using pentodes V_2 and V_4 are employed for phase inversion. When switch S_4 is at the position marked BINDING POST, duodiode V_6 and duo-triode V_7 are connected respectively as a differential rectifier and a balanced cathode follower similar to those described in the early part of this paper.

The high-frequency probe consists of a germanium crystal, two resistors and two capacitors. When switch S_4 is at the position marked PROBE, the balanced cathode follower is connected to the probe for operation at high frequencies. Potentiometer R_s is inserted for panel zero adjustment when the probe is in use, otherwise potentiometer R_{\bullet} is used alone for the same purpose. This arrangement was found necessary in order to avoid initial rush of pointer movement during zero setting. Switch S_3 is used for sense identification. If voltage V_1 lags V_2 , the turning of switch S_3 will cause the meter reading to increase since an additional phase angle is added to voltage V_1 .

During operation, when switch S_1 or S_2 is turned from the position shown in Fig. 7 to the position marked 0° or the position marked DIRECT, a sudden large variation in potential difference may develop across capacitors C_8 and C_4 . This variation in turn causes a heavy transient current to overload the output meter. The operation will be interrupted until the charges on C_3 and C_4 return to the steady-state condition. This condition may take as long as several minutes to establish since the input impedance of the balanced cathode follower is very high, over 100 megohms. The heavy transient current may even damage the output meter in some cases unless an automatic relay circuit is added to the instrument.

The relay circuit embodies a triode V_5 , which normally operates slightly above its cutoff, an R-C circuit of proper time constant and an electromagnetic relay of which the action is controlled by the plate current of the triode. When switch S_1 is turned to the position marked 0° from the position shown in Fig. 7, C_1 is short circuited and C_2 with zero initial charge is connected

across the grid and cathode of triode $V_{\rm b}$. The plate current of $V_{\rm b}$ rises abruptly and the relay in its cathode circuit is energized immediately. Capacitor C_2 charges through R_3 and the potential at the grid of $V_{\rm b}$ decreases exponentially toward zero.

The relay will be deenergized again as soon as the plate current of $V_{\mathfrak{s}}$ decreases to a certain value. The energizing of the relay serves to short-circuit the input terminals of the balanced cathode follower and to disconnect the output meter. The charges on C_3 and C_4 will reach their steady-state condition in a much shorter interval and the pointer of the output meter will stay at its zero position until the steady-state condition has been established if a proper time constant is chosen for R_3C_2 . During the next switching action that includes the turning of switch S_1 to its original position or the position marked DIRECT, or the turning of switch S_2 to the position marked 0° , C_2 will be short-circuited and C_1 will be connected across the grid and cathode of triode V_{5} . The relay operates and the process repeats.

The author wishes to acknowledge that the instrument described was developed for the Advance Electronics Company.

SQUARE-WAVE GENERATOR

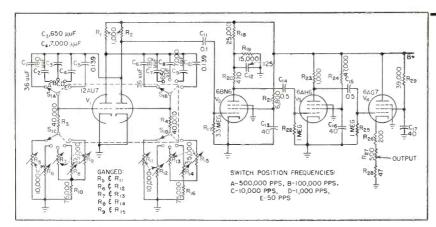


FIG. 1—Circuit diagram for square-wave generator. The clipper stage employs α 6BN6 gated-beam tube

ost square-wave generators are fairly expensive because they employ several cascaded amplifier stages to provide adequate clipping action. With only five tubes used in the complete instrument, a square-wave generator using a 6BN6 gated-beam tube as a clipper delivers a signal with good waveform, fast rise time and wide frequency range. The schematic diagram of such an instrument is shown in Fig. 1.

Circuit Description

Referring to the schematic diagram of Fig. 1, a 12AU7 twin-triode, V_1 , is used as a plate-coupled symmetrical multivibrator. Five separate frequencies of operation are provided with each frequency obtained by changing the grid-plate R-C circuits by means of switch S_1 . The grid resistors are dual potentiometers which can be adjusted to supply exact frequencies on each switch position regardless of coupling-capacitor tolerances.

The nonsinusoidal signal appearing across plate load R_2 is applied through C_{11} to the gated-beam clipper V_2 . Both plate and screen voltages on this tube must be kept low for good clipping.

A square-wave signal appears across plate load R_{20} and is applied through C_{14} to the wide-band volt-

age amplifier V_3 . The amplified square-wave signal is then applied through C_{15} to the cathode-follower output stage V_4 which permits a low-impedance output.

Direct voltages to operate the instrument are supplied by a conventional full-wave rectifier power supply using a 5Y3GT.

Design Features

Small-value plate-load resistors are used in each amplifier stage to provide wide frequency response and to minimize the effect of distributed capacitance.

To amplify a square-wave signal with minimum waveform distortion, an amplifier should be capable of amplifying frequencies without attenuation to at least the tenth harmonic of the square-wave fundamental frequency. Since the highest frequency square-wave signal supplied by this unit is 500,000 pps, the amplifiers must be reasonably flat to better than 5 megacycles. By using small plate-load resistors and care in wiring it has been possible to accomplish this without having to use peaking coils or other means to boost the high-frequency response. This is desirable so that the final square wave will be free of overshoots, ringing and high-frequency phase shift.

At low frequencies, other prob-

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lems are encountered. Since a low-frequency square wave may be considered as a switched d-c signal, the amplifier circuits must be capable of holding a steady level during a period equal to one-half a cycle of the lowest-frequency square wave to be obtained. Towards this end, large R-C time constants are used in coupling between stages and large-capacitance electrolytics are used in filtering and by-pass circuits.

Several requirements dictated the design of the output circuit. In common with all wide-band devices, the output impedance had to be low. At the same time, it was desired to provide control over the output level at minimum cost and without distortion of signal waveform at different levels.

A cathode-follower output circuit was selected and a low-resistance potentiometer, R_{27} , used to control the output level. Even with these precautions, the waveform suffered near the ends of rotation of the control. This was due to the frequency consciousness of a potentiometer type of control. To minimize this effect, fixed resistances, R_{20} and R_{20} , were added to prevent adjusting the control to extreme limits.

In the final circuit, it is possible to adjust the output level over the desired range without waveform distortion from 50 to 500,000 pulses per second.

Choice of Frequencies

Since a square-wave signal is satisfactory for checking an amplifier to at least the tenth harmonic of its fundamental repetition rate,

USING GATED-BEAM TUBE

Output from 50 to 500,000 pulses per second is provided in five steps for checking frequency response, phase shift or transient response. Simple circuit includes a symmetrical multivibrator, gated-beam tube, wide-band amplifier and cathode-follower output

it is not necessary to provide continuous frequency coverage in a square-wave generator. Four or five frequencies, if carefully chosen, are satisfactory for all normal work.

Frequencies permitting the final instrument to be applied with equal versatility to the design of audio amplifiers and pulse and video amplifiers were selected. For these reasons, repetition rates of 50, 1,000, 10,000, 100,000 and 500,000 pps were chosen.

Another model will be made available for wide-band, high-fidelity audio-amplifier work. It will supply signals at 50, 500, 5,000, 10,000 and 20,000 pps.

Output-signal level may be adjusted by means of $R_{\rm sr}$ from approximately 0.8 to 8 volts peak-to-peak. This permits checking not only complete amplifier circuits but also individual stages and attenuator circuits. Output impedance varies from about 50 ohms to approximately 550 ohms, depending on the attenuator setting.

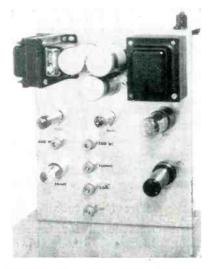
Rise time is good, and for the 500-kc signal is better than 0.07 microsecond.

Applications

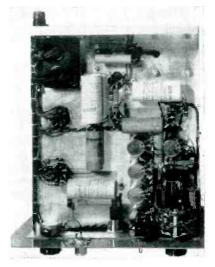
The final instrument may be used for checking amplifiers, peaking or ringing networks, attenuator circuits, phase-shift circuits and pulsing circuits or devices.

For pulse formation, a conventional differentiating circuit may be used in conjunction with the squarewave generator.

When the instrument is used, normal precautions should be followed. Unless necessary to minimize hum pickup, a shielded output



Chassis of square-wave generator, showing dual-potenticmeter adjusting screws for the five output frequencies



Underside of chassis with output potentiometer, output jack, power switch and frequency-selector switch visible

cable should not be used. An open line is preferred to prevent highfrequency signal loss and resultant rounding of the square-wave signal.

Since a direct voltage is present along with the square-wave signal at the output terminal, a blocking capacitor should be connected in series with the hot lead if d-c cannot be tolerated in the circuit to which the generator is connected. The size of this capacitor will be determined by the input impedance of the circuit to be tested and, for this reason, a capacitor was not incorporated as part of the instrument. The capacitor used should be selected to provide an R-C time constant, with the input impedance of the circuit to be tested, at least 25 times greater than the time of one of the lowest-frequency square-wave to be used.

In all cases, connections to and

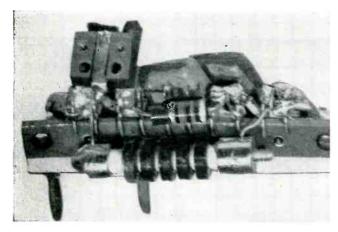
from the circuit tested should be kept as short as practicable. Long leads may introduce sufficient inductance in some amplifier circuits so that resonance at comparatively low frequencies occurs. This results in overshoots and transient oscillation when signals having a short rise time are used. Such distortion of the square-wave signal may lead to incorrect conclusions concerning the circuit tested.

The author wishes to thank Joseph Kaufman, Director of Education, National Radio Institute, for his encouragement and suggestions during the design of this instrument and Mort Massie for his help in securing special samples necessary for the construction of an experimental model.

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TRANSISTOR FREQUENCY



Actual-size photograph of transistor frequency modulator shows simplicity and compactness as compared to vacuum-

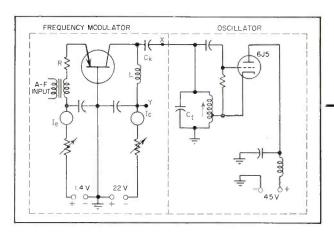


FIG. 1—Transistor frequency modulator provides low-distortion modulation with low d-c and a-f power input requirements.

Signal frequencies up to 1.5 mc may be used

RANSISTORS, since their introduction several years ago, have successfully been applied in many circuits originally designed to operate with vacuum tubes. Experimental work described here shows suitability of transistors for use as frequency modulators.

The control element of the transistor in this application is the emitter as in conventional amplifier circuits. The controlled element is, however, the internal impedance of the collector-base path. Measurements on transistors show that the internal collector impedance changes between broad limits if the voltage and the current are changed slightly between the emitter and base. The measurements further show that instead of the internal collector impedance being purely resistive, it is actually complex for some values of emitter current.

In some transistors the reactive component of the collector impedance changes its value from inductive to capacitive as the emitter current is changed. In other transistors the sign of the internal collector impedance remains the same within the range of emitter currents used.

The wide variation of internal collector impedance is used to produce frequency modulation in an oscillator circuit. A conventional tube circuit is used for the oscillator. The production of frequency modulation by resistance changes

has been proposed before^{1,2}. These methods use the variable internal plate resistance of tubes and some resistance-transforming means for coupling variable reactive elements to an oscillator tank.

Circuit Details

The circuit used for the transistor frequency modulator is somewhat simpler. Figure 1 shows the diagram of the experimental setup. The collector circuit of the transistor is fed with direct current through r-f choke L. A capacitor, C_k , not greater than about 1/10of the oscillator tank capacitor, couples the collector to the oscillator tank. The emitter circuit is biased to class A amplifier service. The audio input is applied to the emitter circuit through series resistance R. This resistance reduces distortion caused by the nonlinear characteristic of the emitter circuit. The transistor cannot develop an audiooutput voltage in the collector-load circuit, even if the collector current is varied, because r-f choke L acts as a short-circuit for the audio output. The audio signal applied to the emitter, however, varies the collector resistance between broad limits at an audio-frequency rate. Coupling capacitor C_k is coupled to the tank through the variable internal collector resistance of the transistor and the ground return. The r-f choke in parallel with the collector represents a high r-f reactance and can be neglected.

The coupling capacitor draws more or less capacitive current from the tank according to the instantaneous value of the transistor internal collector resistance. This produces a frequency modulation. The oscillator tank, however, is loaded through C_k with a variable resistance, which is the internal collector resistance. This variable loading effect produces some amplitude modulation. By proper selection of the oscillator tank capacitor, C_t , which should be of low reactance, the amplitude-modulation effect of the variable collector resistance can be kept below any desired value. A high value of reactance for the tank capacitor is needed, however, to produce a large frequency deviation for a given collector resistance variation. A sound compromise is possible between these contradictory requirements.

Figure 2 shows a typical static linearity curve for a transistor frequency modulator operating at 1.03-mc carrier frequency. During the static deviation measurements the flow of modulation current was simulated by varying the emitter bias-current resistor. Also shown in Fig. 2 is the amplitude variation of the frequency-modulated oscillator tank.

Figure 3A shows the equipment used to test the deviation and static linearity of the circuit. The signal from the oscillator under test is

MODULATOR CIRCUIT

Use of transistor in frequency-modulation circuit saves a-c filament power, reduces size and weight, requires less a-f driving power and reduces d-c power requirements. Distortion in transistor frequency modulator can be kept as low as in tube modulators

By L. L. KOROS and R. F. SCHWARTZ

Engineering Products Department, Advanced Development Section RCA Victor Division Radio Corporation of America Camden, New Jersey

picked up on a communications receiver (with c-w oscillator on) and the receiver output frequency deviations are measured by comparison with a calibrated audio signal generator using Lissajous figures.

The circuit of Fig. 1 was tested for its dynamic performance with a 400-cycle modulating signal. This was done by applying the audio signal to the a-f input terminals and observing the output by means of a discriminator and wave analyzer. Two 6SH7 limiter stages preceding the 6H6 discriminator seemed to give sufficient limiting for all signal levels encountered in this work and assured a linear output response of ± 2.5 volts d-c for $\pm 5,000$ -cps frequency variation with a 0.05-volt rms frequency-modulated input.

A distortion analysis of the modulator was made for three different values of deviation. Figure 3B shows graphically the relationship between the input voltage and the deviation for one of the transistors investigated. The results of the distortion analysis are shown in Table I. The distortion components were read on the wave analyzer at the discriminator output.

A measurement on the audio oscillator indicates that for the maximum deviation setting shown in Table I, and for a resistive load of the same order of magnitude as that represented by the modulator circuit, the output distortion of the audio oscillator was 0.3 percent rms. For output settings corresponding to the lower inputs used, the distortion is so low as to be considered negligible. No correction was used, therefore, on the data of Table I. Figure 3C shows the plot

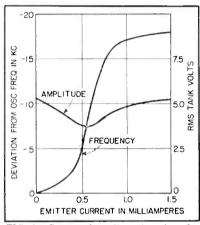


FIG. 2—Curves show linearity of modulation characteristics and effect of modulation on oscillator output amplitude

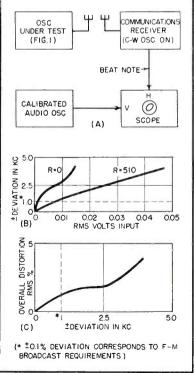


FIG. 3—Equipment used to test transistor frequency modulator is shown in A.

Resulting curves for a typical transistor are given in B and C

of overall distortion versus deviation for the transistor considered here. The audio-frequency response of the transistor modulator was measured and is substantially constant from 20 to 15,000 cps for all of the transistors investigated.

The data given in Table I and Fig. 3C show that for a ± 0.1 -percent deviation of the carrier frequency, which is the upper limit required in the f-m services, the distortion was reasonably low. The interpolated value was only 1 percent rms. Figure 3B shows that the input voltage to the modulator to obtain this deviation was also low. The ± 0.1 -percent frequency deviation was achieved with about 0.008 volt rms input. The input resistance of the emitter circuit with a 510-ohm series resistor R was about 650 ohms. For maximum deviation the input requirement of the modulator with the linearizing resistor was therefore about -40 db (0 level 1 milliwatt). The input without linearizing resistor was -55 db. The input requirement showed fluctuations in the order of 6 db from transistor to transistor.

Internal Collector Resistance

Two equivalent circuits of the modulator are shown in the insert in Fig. 4. Any collector reactance is included in C_s . This form of representation does not follow the real situation because the coupling capacitor has a predetermined value and the changing element is the collector impedance. It is more convenient, however, to separate the resistive and reactive elements into two groups and to consider the modulator as a series or parallel

combination of a capacitor and resistor. The equivalence of the series and parallel representation is expressed by

$$R_s - jX_s = -\frac{R_p jX_p}{R_p - jX_p} \tag{1}$$

$$\frac{X_{\bullet}}{R_{\bullet}} = \frac{R_{p}}{X_{p}} \tag{2}$$

where X_s and X_p are the reactances

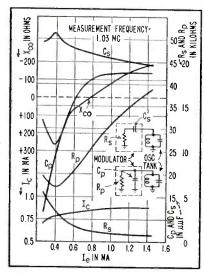


FIG. 4—Curves show relationship between emitter current and other circuit variables for two forms of equivalent circuit

of C_s and C_p respectively at the oscillator frequency.

For practical purposes, the data which can be measured directly by the easiest method are C_p and ΔC_p , the change of C_p . Resistance R_p can be measured by a Q-meter, and C_s and R_s can be computed.

From Eq. 1 and 2

$$C_s = C_p - \frac{X_p^2 + R_p^2}{R_p^2}$$
 (3)

$$R_* = R_p \frac{X_p^2}{X_p^2 + R_p^2} \tag{4}$$

The frequency deviation Δf is expressed by

$$\Delta f = -\frac{f}{2C_t} \Delta C_p \tag{5}$$

If it is found that the calculated $C_s = C_k$, which is the coupling capacitor actually used, the internal collector impedance is purely resistive and its value is identical with R_s . If, however, it is found that C_s is less than C_k it means that the internal collector impedance can be represented by a resistor and capacitor in series. If C_s is greater

than C_k the internal collector impedance can be represented by a series combination of a resistance and inductance.

The technique used to measure C_p , ΔC_p and R_p consists of the determination of the Q of the network formed by the internal collector impedance and the coupling capacitor C_k connected in series. The frequency modulator part of Fig. 1 is coupled to the Q-meter at X and Y. The inductance of the Q-meter, for obtaining the most exact figures, should be chosen lower than recommended for normal use of the instrument at the same frequency, in order to make the Q reading on a convenient range on the meter. If lower inductance is used it is necessary to add capacitors in parallel to the variable capacitor of the Qmeter to tune the circuit to resonance at the investigated frequencies. The techniques of the measurement are as follows: (1) Tune the Q-meter's variable capacitor to maximum Q-reading, before the transistor is inserted in its socket. Read Q_o and C_o . (2) Plug in the transistor and select different emitter current values I_{ϵ} . (3) Tune the Q-meter's variable capacitor to maximum Q-reading again with each I_{ϵ} setting and read the $Q(I_e)$ and the changes on the Qmeter capacitor ΔC_{ν} . From the Q readings for the different emitter currents, R_p is computed as

$$R_p = \frac{Q_o Q(I_e)}{Q_o - Q(I_e)} X_L \tag{6}$$

where X_L is the reactance of the Q-meter coil used. To minimize errors due to distributed capacitances and lead inductances, the transistor socket with the coupling capacitor C_k and the r-f feed choke L was built as a compact unit which was directly plugged into the terminals of the Q-meter. A photograph of this frequency modulator unit is shown.

The curves of Fig. 4 show the values of C_p , R_p , C_* and R_* as functions of the emitter current at 1.03 mc for one transistor. The experimental data are presented in Table II. The values of the collector current I_c are also shown in the same figure. The values of C_p and Q were measured, and the values of R_p , C_* and R_* were calculated from

Eq. 6, 3 and 4. The reactive component of the internal collector impedance is represented by X_{co} . The computation of X_{co} was done with the expression $X_{co} = X_{ck} - X_{s}$ where X_{ck} is the reactance of the coupling capacitor and X_s is the reactance of C, at the r-f carrier frequency. As Fig. 4 shows, the value of the reactive component X_{co} was found to be low compared with R_s. Nevertheless, it had an influence on the frequency deviation. Reactance X_{co} changed in this case from inductive to capacitive values. There were found, however, transistors where X_{eq} was even smaller and the internal collector impedance was, for the main part, resistive. It was also found that the collector resistance and reactance depended upon the electrical history of the transistor. In the investigated cases, however, if the transistor showed amplification properties, the frequency-modulation property also was present.

Figure 4 shows that a large variation of the resistive component of the collector impedance, about 600 to 5,000 ohms, takes place for only a small change of the order of one milliampere in emitter current. Since the emitter resistance is of the order of 100 ohms, only a small input power is required for a relatively large change in collector resistance, as we have previously seen. Such a variation is an interesting feature which makes the transistor especially useful as a frequency modulator.

The transistor frequency modulator may introduce some jittering of the oscillator frequency due to instability of the transistor constants and to temperature changes. If the center frequency of the oscillator is stabilized by means of a discriminator and standard frequency

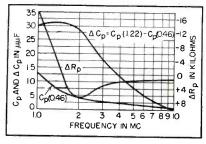


FIG. 5—By careful choice of transistors, operating frequency of frequency modulator can be extended up to 10 mc

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source, as is the standard practice for broadcast transmitters, a considerable part of the effect of transistor instability is eliminated.

Modulating Higher Frequencies

The relatively low output power of the frequency-modulated oscillator cannot be considered as a limitation of the system, because the frequency multiplier stages amplify and multiply the r-f excitation to the necessary power level to excite the f-m power amplifier tube.

To obtain information about the capability of the transistors to modulate oscillators at higher frequencies, several transistors were tried with the previously described Q-meter method at frequencies between 1 and 10 mc. Figure 5 represents typical experimental data. Table III shows the measured values.

The emitter current limits of 0.46 ma to 1.22 ma (with the corresponding collector current units of 0.74 ma to 0.86 ma) were selected to produce high C_p changes around the linear response part of the frequency-modulation characteristic. The R_p values were computed again by Eq. 4. The changes of R_p (ΔR_p) are represented also in Fig. 5. The sign of ΔR_p indicates whether the parallel resistance value is going down with increasing emitter current, in which case ΔR_p is shown as positive, or going up, in which ΔR_p is shown as negative. This increasing or decreasing character depends, in the case of one investigated transistor, upon the applied frequency. In the case of another transistor, R_p decreased with increasing emitter current for all frequencies between 1 and 10 mc.

Figure 5 shows that this transistor unit with the coupling capacitor of 47 $\mu\mu$ f has a reasonably high ΔC_p value, and consequently has frequency modulation capability up to 2.5 mc. The behavior of the transistors is not uniform at higher frequencies. It is too early to say what is the upper frequency limit of the transistor in f-m service. Of course the flexibility of the modulator will be higher if the oscillator frequency can be increased. Otherwise, at least one more frequency multiplier stage must be used. This additional

Table I-Overall Measured Distortion

Input Voltage to Modulator (rms volts)	Deviation	Distortion Co 400-eps mod	Overall Distortion	
	(± cps)	Harmonic	% Fund.	(rms %)
0.0132	1,350	2nd	0.979	
	-, -	3rd	0.769	
		4th	0.419	
		5th		1.32
0.0330	2,850	2nd	1.91	
		3rd	0.800	
		4-th	0.450	
		5th		2.12
0.0160	3,900	2nd	3.62	
		3rd	0.75	
		4th	0.30	
		5th	0.337	3.72

Table II-Measurement of Collector Variables

I _e (ma)	I _c (ma)	Q	C_Q^*	C_p $(\mu\mu f)$	R_p (Kilohms)	C_s $(\mu \mu \mathbf{f})$	R_s (Kilohms)	(Kilohms
		106**	177	0	Before ins	serting	ransistor	
0.46	0.74	55	165	12	7.37	48.4	5.55	+0.09
0.58	0.78	50	155	22	6.12	51.1	3.48	+0.26
0.75	0.81	54	142.5	34.5	7.12	48.2	2.02	+0.08
0.96	0.84	62.5	136	41	9.84	47	1.26	0
1.22	0.86	70	134.5	42.5	13.35	45.7	0.92	-0.09
1.64	0.86	75.5	134.5	42.5	16.98	14.4	0.74	-0.19

^{*} C_Q is the reading of variable capacitor in Q-meter. A fixed capacitor was used parallel to C_Q to tune the 10-microhenry coil to 1.03 mc.

** The first line represents Q_o , the others $Q(I_o)$.

Table III—Measurement of Frequency Response

f (mc)	Coil (µh)	Q (0.46)	Q (1.22)	$rac{C_Q(0.46)^*}{(\mu\mu{ m f})}$	C_Q (1.22) * $(\mu \mu { m f})$	$C_p^{**} (\mu \mu \mathbf{f})$	$\Delta C_p \ (\mu \mu f)$
1.03	10	55	70	165	134.5	12	30.5
2	5	60	39	204	175	4	29
3	1	56	42	261	243	3	18
4	1	71	40	268	255		13
5	1	72	32	126	120	2	б
7.5	1	71	21	418	414	2	4
10	1	59	20	225	225	1	0

^{*} C_Q (0.46) and C_Q (1.22) represent readings of variable capacitor in Q-meter at 0.46 and 1.22-ma emitter currents.

stage might also be necessary for high-power transmitters due to the low oscillator power. As transistor development advances and the similarity between units becomes greater, the reasonable upper frequency limit and oscillator tank voltage limit for the transistor frequency modulator can be determined.

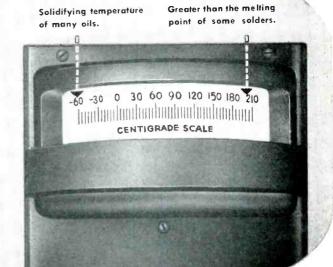
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(2) B. E. Montgomery, Inductively Coupled Frequency Modulator, *Proc. IRE*, p 559, Oct. 1941.

^{**} C_p is capacitor value represented by transistor at 0.16-ma emitter current.

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TUBES AT WORK

Including INDUSTRIAL CONTROL

Edited by RONALD K. JURGEN

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Acoustic Mill-Feed Controller

Pulverized furnaces and boilers draught-fed furnaces and boilers is processed in what are known as tube mills. Coal fed continuously into this type of mill is reduced to the consistency of fine grade powder by the action of cascading steel balls within a rotating tube or drum. Efficiency is almost entirely dependent on the correct amount of raw coal being continuously maintained within the tube.

In fuel processing mills, the rate of feed required is dictated by a number of variable factors. Therefore, to be effective, any method of automatic feed control must be based on the actual, and not estimated, amount of raw coal within the tube or drum.

Controller Apparatus

Since it is not possible to make an assessment of the quantity within the tube based on the amount of coal fed into the mill or its output, the acoustic mill-feed controller has been designed to determine the actual contents of the mill from the noise level of the cascading coal and balls in contact with each other and the walls of the drum. As with the human ear, this determination calls for fine distinctions in noise level taken at a reference point in the immediate vicinity of the drum.

The basic apparatus consists of a

microphone, adequately protected against dust and dirt, placed immediately beneath the drum. The microphone is connected by cable to the main controller unit which is set up at any desired distance from the tube mill.

The output stage of the controller unit comprises a series of pads for lining-up purposes, an amplifier and a high-pass filter. The function of the filter is to segregate the significant high-frequency noises emanating chiefly from inside the drum from the low-frequency noises associated chiefly with the external gearing and drive of the mill itself.

The most significant noise frequencies have been found to exist at about 4 kc. These frequencies give a noise discrimination of about 14 db between the full and empty condition of the tube as compared with about 3 db, which is the best discrimination possible when considering the over-all noise frequency spectrum of the mill.

The output from the amplifierfilter stage of the controller is used to actuate a sensitive relay adjusted to operate in accordance with fluctuations in noise above and below a set figure corresponding to the optimum amount of coal in the tube. The relay serves to provide, through suitable heating coils and

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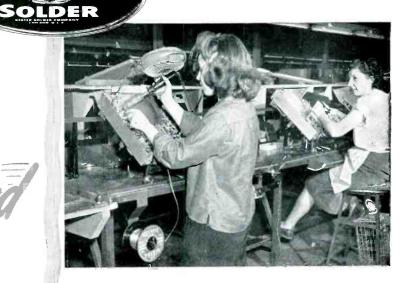
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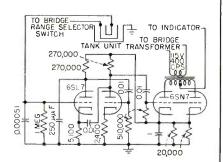
KESTER SOLDER

THE FRONT COVER

HE COVER photograph this month shows a production tank unit for the Simmonds electronic fuel gage in the process of being tested to see if tank unit electrical capacitance compares favorably with quality control requirements. The inspection is being made on a Master Laboratory capacitance bridge.

The capacitance bridge network consists of a Campbell Shackelton shielded ratio box, a Hewlett-Packard a-f signal generator to supply alternating voltage and a detector which works through an amplifier and a 1,000-cycle selective filter or straight through for other frequencies. Leeds and Northrup single-range potentiometer indicators are also used.

The Pacitron Gage system consists of a tank unit of variable capacitance, an amplifier as shown in the accompanying schematic, a bridge calibrator and a



motor-driven indicator. In the system, the capacity of the tank unit is constantly compared with a fixed-capacitance reference capacitor.

The voltage resulting from a difference in capacitance is amplified and applied to a motor-driven potentiometer which restores the bridge to balance. The quantity of fuel, as sensed by the tank unit, determines the potentiometer wiper position. An indicator pointer attached to the potentiometer indicates the quantity of fuel,

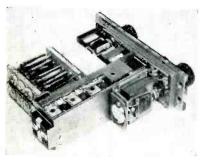
relay trains, a means of altering the speed of the feed motor or, alternatively, for manual control purposes, of actuating a suitable gage to afford visual indication of

the contents of the tube mill.

The equipment is manufactured by Standard Telephones and Cables Limited in Aldwych, London, England.

Subminiature Low-Frequency Receiver

A SUBMINIATURE radio receiver for aircraft use, continuously tunable from 190 to 550 kc and utilizing a 135-kc i-f amplifier, has



New subminiature receiver chassis ready for encasement and hermetic sealing in nitrogen

been developed by the National Bureau of Standards. This assembly occupies about 55 cubic inches and is the functional equivalent of a World War II unit more than five times as large.

The new 12-tube receiver, in conjunction with a previous high-frequency project (a 60-mc, 11-tube i-f amplifier assembly), effectively brackets the communication spectrum. The receiver is used to keep aircraft on course. It has a sensitivity of 5 μv for 6 db signal-tonoise ratio and a power output of 100 milliwatts.

The 12 tubes provide two tuned r-f amplifier stages, a mixer, a local

oscillator, two 135-kc i-f amplifier stages with a bandwidth of about 2 kc, a diode detector, an avc diode, a beat-frequency oscillator, an audio amplifier stage and a pushpull-parallel power output stage. All stages operate with 26 volts d-c on heaters, screens, and plates. Under these conditions of operation, four subminiature audio power output tubes are required for adequate power output.

To facilitate mass production, seven detachable subassemblies, each of which can be built independently, are employed in the receiver. This also permits easier servicing. The subassemblies are fastened to one another and to the front panel, which takes the place of a chassis.

Printed circuits are used to a considerable extent. They are of value both in conserving space and for economical mass production. The wiring is printed on steatite or silicone-impregnated fiber-glass.

Hermetic Sealing

Hermetic sealing of the entire unit affords protection against moisture and contamination. It also permits elimination of protective coatings for the individual components, which saves space. A soldered copper band seals the housing to the front panel. The band can be removed for repair purposes with a key of the type used for opening coffee cans. The air in the unit is replaced with nitrogen before sealing to prevent oxidation.

The tuning of the receiver over its wide operating frequency range was a difficult engineering problem. A straight-line tuning characteristic (frequency proportional to control angle) was attained in the limited space allowable by the use of a variable-pitch screw to drive the slugs in and out of the r-f coils.

The tuning slugs are made of the newly developed ferrite base materials, more stable at high temperatures than powdered-iron cores. It was expected that production control of these new materials might present a difficulty; mechanical means were therefore provided in the tuner to compensate for non-uniformity of core material.

The i-f transformers have an over-all size of ½ by ½ by 1% inches; (continued on page 156)

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THE ELECTRON ART

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Checking Quality of Aluminized Screens

BY HOWARD J. EVANS

Tube Department Radio Corporation of America Harrison, New Jersey

To avoid waste manufacturing steps, two methods have been developed to test aluminized picture screens before the gun structures are sealed into the bulbs. These developments were necessary because the standard test on non-aluminized screens cannot be used on aluminized ones.

Test Procedure

In the test on nonaluminized screens, after the phosphor has been deposited on the faceplate, the bulb is partially evacuated, and a high-frequency spark coil is moved over the faceplate. The resultant intense electric fields inside the bulb cause high-velocity electrons to strike the phosphor, exciting it so that actual operation of the finished tube is simulated and the quality of screen luminescence can be visually examined.

However, in bulbs which use an aluminized coating on the back of the phosphor to prevent ion-spot formation, the aluminum acts as a radio-frequency shield which prevents any outside electric fields from penetrating into the bulb enclosure. Formerly, therefore, the screens of aluminized picture tubes could not be fully tested until the tube was completed and examined in actual operation. Each defective screen, consequently, resulted in virtually a total loss, for the cost of

salvaging the good parts of the tube almost equalled the original cost.

Two Methods Devised

The two methods devised to prevent this waste permit a check of the quality of aluminized screens earlier in the manufacturing process. One, the radio-frequency method, uses a high-frequency oscillator to produce an electric field which impels high-speed electrons against the phosphor. The second, the direct-current method, accelerates the excitatory electrons by

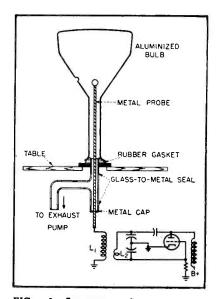


FIG. 1—Apparatus for examining screens of aluminized picture tubes by r-f method

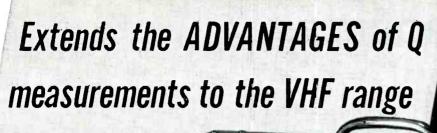
means of direct voltage. This d-c method, although not now used in practice, will be described because it conceivably might be preferable if the presence of a strong radio-frequency field were undesirable.

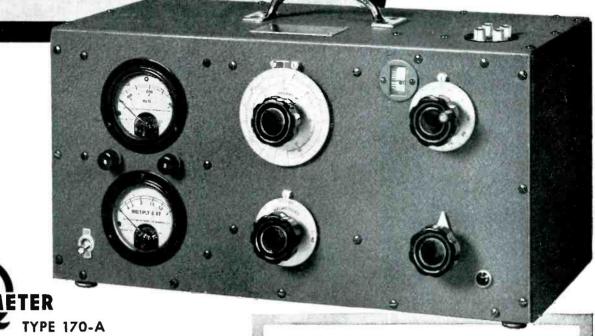
Radio-Frequency Method

A representation of the r-f method is shown in Fig. 1. The bulb to be examined is connected in an upright position to the intake port of a vacuum system and the bulb is evacuated. The flanged portion of the bulb neck presses against the rubber gasket with sufficient force to form a gas-tight seal. When the final vacuum is established, this force is equal to the weight of the bulb plus the product of the atmospheric pressure and the area of the bulb flange. The rubber gasket is restrained from collapsing inward by the metal cylinder which is placed inside the gasket.

The electrical energy which causes screen excitation enters the evacuated bulb through the probe. When the probe is driven positive or negative, an electric field set up between the probe and the aluminized coating accelerates the free electrons or ions. The ions have a mean free path of the order of 1 centimeter when the pressure is 5 microns. This path is long compared with the radius of the neck, so that few ionization collisions occur in the neck region. However, the mean free path is short compared with the probe-to-screen distance, so that sufficient ionization collisions occur in the cone region to maintain the supply of ions and electrons. On the positive half of the r-f cycle the positive ions are repelled toward the screen. Since there is no electrical connection to the screen, these positive charges cause the screen to assume a positive potential with respect to ground. On the negative half of the cycle the probe becomes negative, and electrons and negative ions strike the aluminum coating with a maximum energy which, when measured in terms of electron volts, exceeds the maximum voltage between probe and ground. This energy is sufficient to cause the electrons to penetrate the alumi-

(continued on page 184)





Since the first use of tuned circuits, the ratio of their reactance to their losses at resonance has been an important factor of merit. This ratio, known as Q, specifies the band-width, cut-off rate and potential gain of the tuned circuit when used in an amplifier. The Q Meter is a standard of the industry because it quickly and accurately measures this fundamental, unchanging quantity.

Freq. 30-200 mc.

Frequency ranges, tuned circuit designs and circuit applications have changed. To keep pace with these changing requirements, Q Meters have been introduced for different frequency ranges and with other design changes. The Type 170-A Q Meter is one of the results of our continuous program of developing precision equipment to meet the needs of laboratory and industry.

Write for complete information.

DESIGNERS AND MANUFACTURERS OF THE Q METER • QX CHECKER FREQUENCY MODULATED SIGNAL GENERATOR • BEAT FREQUENCY GENERATOR AND OTHER DIRECT READING INSTRUMENTS

MEASURES

- . Q of VHF Coils.
- Effective Inductance of Coils.
- Losses in Condensers and Dielectrics.
- Characteristics of VHF Antennas.
- Distributed Capacity of Coils.

SPECIFICATIONS:

Oscillator Frequency Range: Continuously variable from 30 mc. to 200 mc. in 3 ranges.

Frequency Calibration Accuracy: # 1%.

Range of Q Measurements: Directly Calibrated—80 to 300. By use of Multiplier—80 to 1200.

Accuracy of Q Measurements: Directly Calibrated Scale ± 10% to 100 mc.

Accuracy decreases with frequency above 100 mc.

Capacitance Calibration of Q Condenser:
Range 11 to 60 mmfd. with 1 mmfd. divisions.
Micrometer Vernier marked with 100 divisions.
Accuracy: 1% or 0.5 mmfd., whichever is greater.

Inductance Range: 0.05 microhenry to 1.3 microhenry.

Power Supply: 110–120 volts, 50–60 cycles. Also 220–240 volts, 50–60 cycles. Power Consumption—50 watts.



NEW PRODUCTS

Edited by WILLIAM P. O'BRIEN

Military Electronics Program Extends Miniaturization Trend . . .

Noncritical Materials Stand Out in Design for Civilian Use . . .

Available Manufacturers' Catalogs Are Listed



Tiny Electronic Amplifier

BENDIX RADIO DIVISION OF BENDIX AVIATION CORP., Baltimore 4, Md., has developed a tiny electronic amplifier for miniaturized airborne radar equipment. It is a band-pass type amplifier with a rated electrical gain of 130 db. Weighing less than 2 ounces and requiring only 35 component parts, its cost in large production quantities should be small enough to make feasible replacement of the entire amplifier in case of tube or component failure and because of its plug-in design replacement can be made in a matter of seconds. Illustrated is the unit as compared to a conventional band-pass amplifier as used in radar equipment,



Polyphase Reproducer

AUDAK Co., 500 Fifth Ave., New York 18, N. Y., has announced the Audax L-6-G, a new polyphase reproducer. This single magnetic

unit will play any and all lateral recordings at speeds of $33\frac{1}{3}$, 45 or 78 rpm. A special connector is available which permits the unit to be plugged into the Garrard changer arm. Point pressure is 8 grams for all discs. Output is approximately 20 mv. Response is from 20 to over 10,000 cps. The sapphire or diamond stylus is easily replaceable.



High-Temperature Variable Resistor

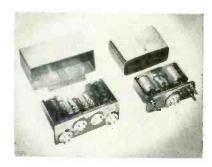
CHICAGO TELEPHONE SUPPLY CORP., Elkhart, Ill. Type 65 miniaturized variable composition resistor has extremely high temperature and humidity stability. It is specially designed for use in all types of military communications.



Mounting Bases

T. R. Finn & Co., Inc., 333 Jackson Ave., Bronx 54, N. Y. Finnflex mounting bases, designed to protect airborne equipment such as aircraft radios, radar, loran and similar electronic apparatus, exceed Army-Navy E-19 drop test require-

ments. They are designed to meet dimensional and metal requirements of JAN C-172A specifications. Load range is from 6 to 80 lb. The unit illustrated is carried by four metallic spring-type isolators, which provide protection by maintaining their high vibrationisolating efficiency from $-80~\mathrm{F}$ to $+250~\mathrm{F}$.



Sealed Multiple Relays

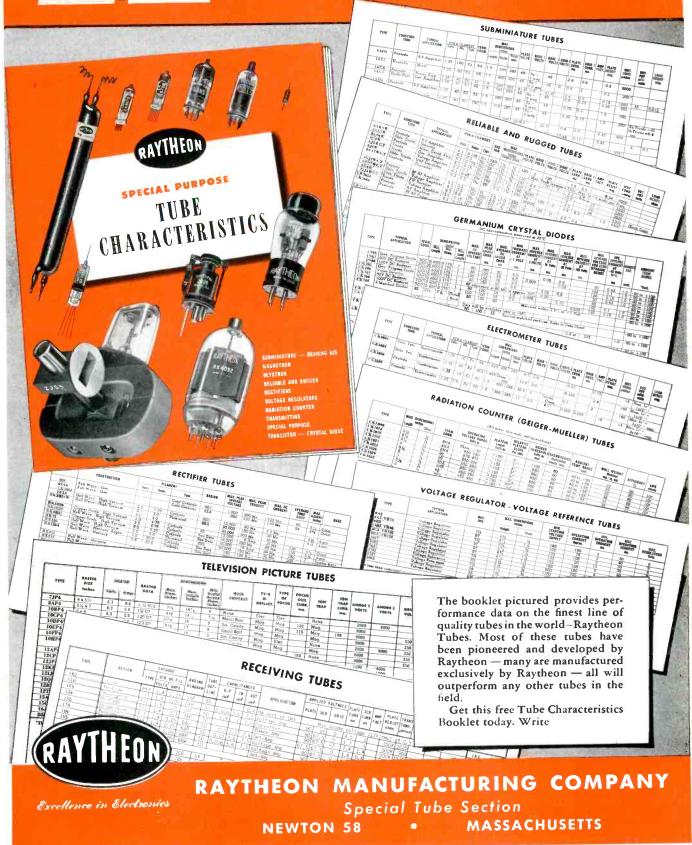
Potter and Brumfield, Princeton, Indiana, has developed a line of 2-in-a-can and 4-in-a-can sealed relays for use in military airborne equipment. The 2-in-1 assembly has two-ampere palladium contacts on each relay. It has overall dimensions of 3 in. \times 1½ in. \times 2 in. high and weighs 6 oz. The 4-in-1 assembly has 5-ampere silver contacts. It measures $3\frac{\pi}{8}$ in. \times 1½ in. \times 2 in. high and weighs 13 oz. The relays are thoroughly desiccated and hermetically sealed in inert gas.



Sealed Stand-Off

ERIE RESISTOR CORP., 644 W. 12th St., Erie, Pa. Basic design of the style 326 stand-off Ceramicon for whf and uhf applications provides, in a hermetically sealed case, a bypass-to-ground through the shortest possible path. Post terminal provides a sturdy tie point for several connections, and is made essentially to match tube socket terminal height in the interest of

Write For Your Free Copy Of This Complete RAYTHEON TUBE GUIDE



maintaining uniform short leads. Designed for threaded mounting installation, it is available in standard capacitance values in $\mu\mu$ f: 10, 33, 47, 68, 82, 100, 680, 1,000 and 1,500. Voltage rating is 500 v d-c.

Miniature Signal Generator

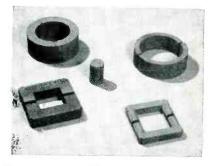
RADIO CITY PRODUCTS Co., INC., 152 W. 25th St., New York 1, N. Y. Model 730 pocket-size, signal generator combines a real f-m signal generator, an r-f signal generator, and an audio oscillator. The four f-m fixed frequencies are modulated to give ample sweep for use with ratio detectors. Alignment can be quickly made at the ends of the band at the 88 and 108-mc switch position. Correspondingly, the 10.7 and 9.1-mc switch positions permit alignment of the i-f sections. The ends of the broadcast band can be readily aligned at the 1,500-kc and 550-kc switch positions. Intermediate-frequency alignment is provided by fixed frequency switch positions of 456 kc and 465 kc. An audio outlet terminal is provided for supplying a 400-cycle signal externally. Weight of the instrument is 2 lb and price is \$32.95.



Air-Damped Mountings

THE BARRY CORP., 700 Pleasant St., Watertown 72, Mass. Series 6475 Barrymounts provide effective vibration isolation with all the ad-

vantages of air damping. Equipment supported on the mountings is raised only $\frac{1}{2}$ in. above the mounting surface. Unit mounts are 1 in. in diameter and have an overall height of 1 in. under minimum rated load. Load ratings range from 0.3 to 3.0 pounds per mount. The center stud is tapped to a depth of $\frac{1}{4}$ in. with an 8-32 thread.



Noncritical Ferrites

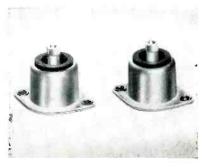
FERROXCUBE CORP. OF AMERICA, 50 E. 41st St., New York 17, N. Y., has announced availability of high-permeability ferrite parts using noncritical materials. Transformer cores, deflection yoke cores, antenna cores, and permeability tuning cores for television and electronics are available in types 3 and 3C materials, which are nickel-free. Complete technical information is contained in engineering bulletin FC-5101.



Record-Reproduce Head

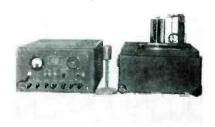
THE BRUSH DEVELOPMENT Co., 3405 Perkins Ave., Cleveland 14, Ohio, has announced the BK-1500 series multichannel (available from 3 to 14 channels) record-reproduce head. Design features include balanced magnetic construction with

gaps at front and rear. All gaps are in perfect alignment and individual channels are cast into one integral block of specially selected synthetic resin. Complete assembly is nonmicrophonic and impervious to moisture. The head has a channel width of 0.050 in. with center-to-center spacing of 0.125 in.; gap is 0.0005 in. Maximum output level at 1,000 cps and 7.5 in. per second tape speed is 0.002 volt rms.



Shock and Vibration Isolators

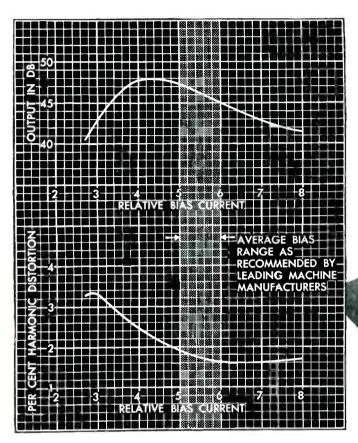
THE BARRY CORP., 700 Pleasant St., Watertown 72, Mass. Series 6465 Barrymounts are designed to meet requirements for smaller isolators to be used with miniaturized equipment. The mountings employ air damping to minimize the shock of aircraft landing and taxiing and to limit excursion so that there is no snubber contact, even at resonance. Load ratings of bases using these mounts are 0.1 to 3.0 pounds per mount. The isolators are 1 in. in diameter and 132 in. high under maximum rated load. The attachment stud has an axial hole ½ in. deep, tapped with 6-32 thread.



Vibration Analyzer

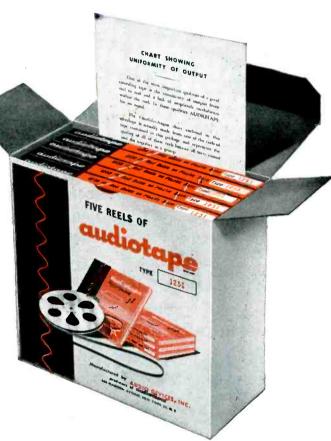
KAY ELECTRIC Co., Pine Brook, N. J. The Vibralyzer is applicable to the $$^{(Continued\ on\ p\ 220)}$$

For MAXIMUM OUTPUT with MINIMUM DISTORTION



you can't beat





And there's PROOF of UNEQUALLED UNIFORMITY in Every Package

You get an Esterline-Angus output curve in every fivereel package of plastic-base Audiotape. This curve, made from one of the reels in that package, actually measures the output characteristics of all five reels, since they are all slit from the same roll after coating. Now you can see, as well as hear, the exceptionally high output uniformity that you get only in Audiotape. What's more, every 1250-ft. and 2500-ft. reel is guaranteed FREE FROM SPLICES!

• As all professional recordists know, the proper operation of any tape recorder involves a compromise between decibels and distortion. And Audiotape has been especially formulated with this important relationship in mind—to give you higher output (and thus better signal to noise ratio) and lower distortion in the normal bias range of all machines.

Test it. Compare it with any other tape. Plot your own output vs distortion curves, similar to the ones shown above. You'll find that in the useful, low-distortion bias range, Audiotape combines maximum output with maximum fidelity and freedom from distortion.

That's just one of many reasons why more and more professional recordists are specifying Audiotape for their most exacting magnetic recording requirements. Remember — Audiotape is made by audio engineers, for audio engineers. It speaks for itself.

AUDIO DEVICES, Inc.

444 MADISON AVE., NEW YORK 22, N. Y.

Export Dept.: 13 East 40th St., New York 16, N. Y., Cables "ARLAB"

audiodiscs audiotape audiofilm audiopoints

NEWS OF THE INDUSTRY

Edited by WILLIAM P. O'BRIEN

DPA Electronics Production Board Appointed

EDMUND T. MORRIS, JR., formerly with Westinghouse Electric Corp. in engineering and managerial capacities, has been appointed to head the Defense Production Author-



E. T. Morris, Jr.

ity's Electronics Production Board. Serving with him are: Capt. F. R. Furth, director of Naval Research Laboratory, Washington, D. C.; Harry A. Ehle, vice-president of International Resistance Corp., Philadelphia, Pa.; John G. Daley, chief of the Electronics Division, National Production Authority; M. W. Boyer, general manager of the Atomic Energy Commission; Don

G. Mitchell, president of Sylvania Electric Products, Inc., New York, N. Y.; and C. W. Middleton, vice-chairman for production of the Munitions Board.

The Electronics Board has the primary responsibility in DPA of assuring that electronics equipment is produced in sufficient quantities and on time to meet requirements of defense and essential civilian programs. Electronics is defined as including not only radio, tv and radar but also extensive application in other fields such as electric motors, fluorescent lighting, automobiles, business machines, fire control and industrial control devices, test and measuring equipment and communications equipment, as well as being an essential component of many basic weapons.

Britain's 1951 Radio Convention

THE British Institution of Radio Engineers' third convention will be held this year during the Festival of Britain. Six separate sessions, to be held at intervals staggered over the period of the Festival, will cover almost the whole field of radio and electronic engineering.

Topic, time and place scheduled for each session are as follows:

Session 1—Electronic Instrumentation in Nucleonics—July 3 and 4, University College, London.

Session 2—Valve Technology and Manufacture—July 5 and 6, University College, London.

Session 3—Radio-Communication and Broadcasting—July 24 and 25, University College, Southampton.

Session 4—Radio Aids to Navigation—July 26 and 27, University College, Southampton.

Session 5—Television Engineering—August 21 to 24, King's College, Cambridge.

Session 6—Audio Frequency Engineering—Sept. 4 to 6, The Richmond Hall, Earls Court.

Amateur Radio to Aid Civil Defense

SUPPLEMENTING commercial communications facilities, the New York State Civil Defense Commission is forming a state-wide amateur radio system to be used for operations in atomic attacks when power or wire communications may be disrupted.

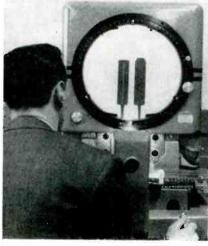
Colonel Lawrence Wilkinson, acting chairman and director of the Commission, has issued a bulletin to the 104 county and city C-D directors, urging them to enlist the support of qualified radio amateurs and their equipment for communications. Under the New York State Emergency Defense Act the State C-D Commission is empowered to control all communication facilities and personnel in event of an attack and during drills and tests prior and subsequent to an attack.

Control will be exercised through the enrollment of operators as members of the New York State Civil Defense Amateur Radio Service. This service will function under the policies, plans and instructions of the commission, of which Vincent T. Kenney is coordinator at the State C-D headquarters in New York City.

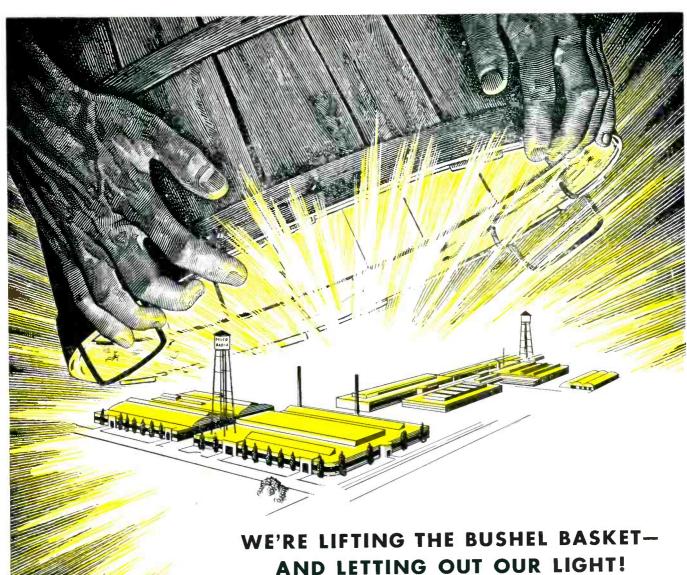
All local C-D directors in the

PRODUCING GERMANIUM DIODES





Girls use microscopes while welding whiskers to pin assemblies at GE's new germanium products plant in Clyde, N. Y. The new plant has a potential capacity of 12 million diodes annually. Photo at right shows a new contour projector in use at the plant. Here the whisker and pellet assemblies are inspected at a magnification of 20 times



AND LETTING OUT OUR LIGHT!

Probably you didn't know it because we haven't told you. BUT-for the past five years we've been building car and truck radios for General Motors. In 1950 alone we built nearly 2,000,000 radios for practically all types of vehicles. Our production tops that of any other car radio manufacturer.

Since we stopped making radio and electronic equipment for the armed forces during World War II, we've concentrated on making our facilities the largest and our products the finest in the industry. Unlike most others, we're not just an assembly plant. We make practically all the parts that go into our products. Our design and research engineers are tops in their field. Our laboratory and production equipment is the finest that can be obtained.

With our vast experience, our facilities, engineering know-how, our productive manpower bigger and better than ever before, we believe we can be of service to our country. We are ready to go to work immediately for national defense and count it a privilege to volunreer for service.

DIVISION, GENERAL MOTORS CORPORATION KOKOMO, INDIANA

State have been informed that a trained nucleus of operating personnel can be obtained from the New York members of the Amateur Radio Emergency Corps, an affiliate of the American Radio Relay League.

Powdered Iron Core Dimensions Standardized

ALL of the types of cores made from iron powder and commonly in use by the radio and television industry have been listed in a new Standard just released by the Metal Powder Association. The Standard was prepared by the Electronic Core Subcommittee of the Metal Powder Association Standards Committee to meet a long felt need in the pulverized core industry for standardization of dimensions and tolerances.

The Standard, designated 11–51-T, defines the terms commonly associated with cores made from powdered magnetic materials and specifies the preferred dimensions of standard sizes and shapes. Preferred dimensions are listed in detail for plain iron cores, insert iron cores, threaded iron cores, tuning cores and sleeve iron cores. In addition to diameters and lengths, data is provided for concentricity, screw-driver slot dimensions, hexagonal hole sizes as well as threaded spring type and spaded insert sizes.

Developed after careful consideration by the members of the Electronic Core Subcommittee, it reflects, in substance, the practices established by the RTMA. Members of the Electronic Core Subcommittee include the following companies: Empire Coil Co. of New Rochelle, N. Y.; Magnetic Core Corp., of Ossining, N. Y.; National Moldite Co. of Hillside, N. J.; Powdered Metal Products Corp. of Franklin Park, Ill.; Pyroferric Co. of New York, N. Y.; Radio Cores, Inc., of Oak Lawn, Ill.; Speer Resistor Corp. of St. Marys, Pa.; and Stackpole Carbon Co. of St. Marys,

William E. Cairnes, president of Radio Cores, Inc., is chairman of the subcommittee.

Copies of the new Standard and other Standards published by the

MEETINGS

Aug. 15-18: 1951 APCO Conference, Everglades Hotel, Miami, Florida.

Aug. 20-23: AIEE Pacific General Meeting, Multnomah Hotel, Portland, Oregon.

Aug. 22-24: Seventh Annual Pacific Electronic Exhibit and West Coast Annual IRE Convention, San Francisco Civic Auditorium, San Francisco, Calif.

Aug. 28-Sept. 8; Eighteenth British National Radio Show, Earls Court, London, England.

SEPT. 10-14: Sixth National Instrument Conference and Ex-

hibit, sponsored by Instrument Society of America, San Houston Coliseum, Houston, Texas.

OCT. 8-10: AIEE Conference on Aircraft Equipment, Hollywood Roosevelt Hotel, Los Angeles, Calif.

OCT. 22-24: 1951 National Electronics Conference, Edgewater Beach Hotel, Chicago, Ill.

OCT. 22-26: AIEE Fall General Meeting, Hotel Cleveland, Cleveland, Ohio.

Nov. 12-15: NEMA Convention, Haddon Hall, Atlantic City, N. J.

Metal Powder Association may be obtained by writing to the Association at 420 Lexington Ave., New York 17, N. Y., and enclosing 25 cents per copy.

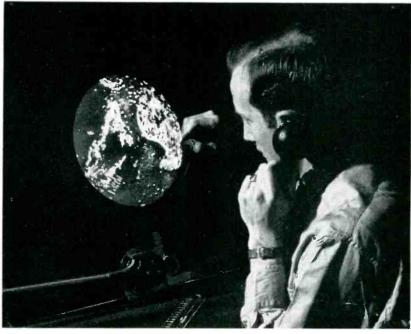
Harbor-Control Radar Tested

PORT of New York Authority, operated jointly by the states of New Jersey and New York, has inaugurated an experimental program to test the feasibility of harbor-control radar as an aid to pilotage during foggy weather.

Shore-based radar installations are now in regular service at Liverpool, England, and Long Beach, California. They are operated experimentally in the harbors at San Francisco and Baltimore. Other installations are under construction in Le Havre, France, and Southampton, England.

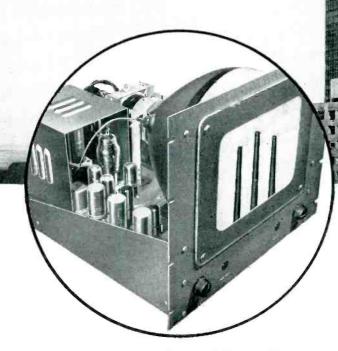
Although no figures are available at present to show the savings that might result with this navigational aid, it is known, for example, that the fixed cost of delaying a 10,000-

(Continued on page 238)



Scope presentation of the New Jersey-New York harbor from Fort Wadsworth, Staten Island, overlooking the Narrows. Ships, islands, buoys and shore lines between Ambrose Light and the Battery are clearly visible

Sylvania Tubes help keep U. N. Building free from smoke



Showing chassis of "Vericon" Picture Monitor equipped with Sylvania tubes. These monitors are finding a wide variety of uses today in industries and institutions. They were designed to the exacting requirements of the Remington Rand "Vericon" System and built by Television Utilities Corp.



The problem of smoke at New York's beautiful U. N. Area is being solved with the help of Remington Rand's "Vericon" Industrial Television System equipped with Sylvania Tubes.

In reaching a satisfactory solution, much credit goes to the Consolidated Edison Company, located near by, for their all-out cooperation. Among other preventive measures, this company installed and focused a set of Remington Rand "Vericon" TV cameras on their towering stacks.

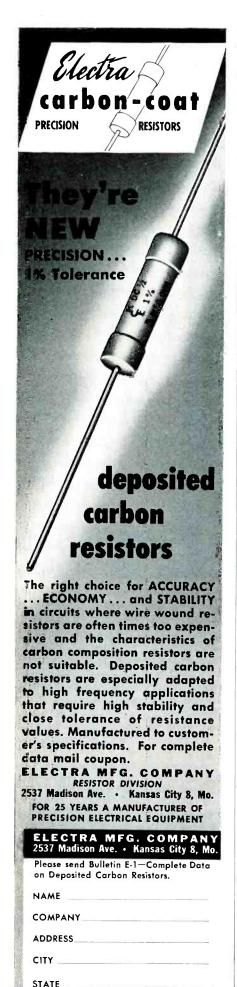
These cameras are hooked up to five strategically located viewing monitors—(soundless TV sets)—made by Television Utilities Corp. and equipped with Sylvania picture tubes and receiving tubes.

Every day, these monitors are in operation. If at any time smoke should appear, Consolidated Edison observers on watch immediately operate special controls to clear up the situation.

Writes Mr. A. E. Siegel, President of Television Utilities: "Our monitors have been running all day long without stop for more than three months. All are equipped with Sylvania radio and picture tubes. We are wondering how long the Sylvania tubes will continue to take this kind of abuse without talking back."

The above is another interesting record of the durability and excellent performance of Sylvania tubes. Let us tell you something about the reasons behind this quality. For full technical data about any types of Sylvania receiving, transmitting, or picture tubes write today to: Sylvania Electric Products Inc., Dept. R-1107, Emporium, Pa. Sylvania Representatives are also located in all foreign countries.

TELEVISION PICTURE TUBES; RADIO TUBES; ELECTRONIC PRODUCTS; ELECTRONIC TEST EQUIPMENT; FLUORESCENT TUBES, FIXTURES, SIGN TUBING, WIRING DEVICES; LIGHT BULBS; PHOTOLAMPS; TELEVISION SETS



NEW BOOKS

Piezoelectric Crystals and Their Application to Ultrasonics

By Warren P. Mason. D. Van Nostrand Co., Inc., New York, 1950, 508 pages, \$7.50.

THE RECENT book by Dr. Mason constitutes a noteworthy addition to his many publications and contributions in this highly specialized field. Not intended for swift or easy reading, this work will become a reference to workers in the field, as opposed to a descriptive treatment for those wishing an elementary survey.

In the opening chapters are developed the basic expressions for the elastic, piezoelectric, pyroelectric, and dielectric relations of sol-

RELEASED THIS MONTH

Electronic Motor and Welder Controls; G. M. Chute; McGraw-Hill; \$6.50.

Elements of Television Systems; G. E. Anner; Prentice-Hall; \$10.35.

Synthesis of Electronic Computing and Control Circuits; Harvard University Press; \$8.00.

ids. Coupled with a concise treatment in the appendix, wherein these relations are erected in tensor form, the development leads to the establishment and individual listing of the fundamental mathematical forms of the seven groups and thirty-two classes into which all crystals fall, as determined by their physical properties and symmetry properties.

The major part of the book summarizes the work and findings of the Bell Laboratories resulting from their many years of significant research in this field. The properties of quartz and rochelle salt are described, but main emphasis is given to such new crystal developments as dipotassium tartrate (DKT), ethylene diamene tartrate (EDT), ammonium dihydrogen phosphate (ADP), potassium dihy
(continued on page 254)

FLECTOR

SELF-LOCKING NUTS



TIGHT

171,360,000

AFTER

VIBRATIONS AT 4000 CYCLES PER MINUTE...

Plain nuts with lock washers loosened after only ONE HOUR of 4000-cycle-perminute operation on the vibrator of a concrete block machine at the plant of the Bethayres Concrete Block Co., Bethayres, Pa.

When FLEXLOC Self-Locking Nuts were installed, they were still tight when the machine was torn down for rebuilding after 6 weeks operation—at 4000 C.P.M., 17 hours a day, 7 days a week!

If you have an application where nuts loosen or back off, try FLEXLOC, the one-piece, all-metal STOP- and LOCK-NUT "that won't work loose."

Send for Bulletin 619-A today.

-SPS-

STANDARD PRESSED STEEL CO.

JENKINTOWN 10, PENNSYLVANIA



When you are faced with specifications that place impossible requirements on dynamotors or small DC motors, according to World War II standards, take advantage of recently developed improvements in high temperature and high altitude techniques by simply outlining your requirements to Bendix. Model units exactly meeting your performance specifications will be developed and tested for pre-production use—production units will then follow in accordance with your manufacturing schedule.

DYNAMOTORS

Regular • Multiple output • Special purpose

DC MOTORS

1/100 hp—1/2 hp • Continuous and Intermittent Duty
DC Servos and special motors

SPECIALIZED DYNAMOTORS

DC MOTORS

RED BANK DIVISION OF BENDIX AVIATION CORPORATION
RED BANK, NEW JERSEY

Export Sales: Bendix International Division, 72 Fifth Avenue, New York 11, N.Y.

Write for this colorful and informative book —it's free. You'll find it loaded with facts and figures about all types of dynamotors.







CORPORATION

1055 NEPPERHAN AVENUE, YONKERS 3, NEW YORK

Backtalk

This department is operated as an open forum where our readers may discuss problems of the electronics industry or comment upon articles which ELECTRONICS has published.

Familiar Music

DEAR SIRS:

I HAVE READ with considerable interest Robert M. Strassner's article on Gas-Diode Electronic Organs in ELECTRONICS for January, 1951.

You undoubtedly will be interested to hear that the principle disclosed in the article is by no means original with Strassner but has been developed by me approximately 17 years ago, together with Francis R. O'Leary my assistant at that time. In this connection, your attention is directed to O'Leary patent 1,986,599 and Langer patent 2,252,189, copies of which accompany this letter.

To complete the references appearing at the end of Strassner's article, I would like to add the following complete list of my pertinent patents which resulted from my experimental work on gas-diode electronic musical instruments extending over the period of nearly a quarter of a century:

1,832,402	2,044,360
1,937,389	2,247,728
1,993,890	2,252,189
2,017,542	2,365,566
2,035,2 38	2,365,567
2,039,651	2,365,568
2,040,439	2,403,664
	NICHOLAS LANGER
	New York, N. Y.

Medical Electronics

DEAR SIRS:

FOR SOME TIME I have felt the need for an organization to bring together those of us who are working in the field of medical electronics. This was further confirmed by the (Continued on page 268)





DYNA-MYKE

The Dyna-Myke Model 129-B is a precision, high speed, dynamic micrometer using linear differential transformers as the sensing element. It measures and provides for recording such phenomena as force, torque, strain, vibration, acceleration, temperature, pressure, thickness, surface finish, etc., with a linear frequency response of DC to 1000 cps. Direct displacements are measured in five ranges from $\pm .1$ inch to ± 10 micro inches. On standard magnetic recorders a sensitivity of 1 micro inch per millimeter is available. A toggle switch converts the Dyna-Myke to a high frequency, high sensitivity strain gage indicator. The output is used to drive any type of magnetic, null balance or galvanometer recorder—or the DC or modulated carrier may be viewed on an Oscilloscope. Selsyn motors may be driven for remote indication or control. Request Technical Bulletin 129-B for full details.



The Dyna-Meter Model 144, when used with the Dyna-Myke, indicates by neon lights the peak amplitude of transients as fast as 1 millisecond. This indication may be instantaneous or a memory feature may be used to maintain the reading until reset. Built-in power relays provide on-or-off control to any plus or minus limits established by the Dyna-Myke. The combination of the Dyna-Myke and the Dyna-Meter offers many applications to industrial processes resulting in the elimination of scrap at the source. Uses in connection with machine tool operations are particularly impressive. Request Dyna-Meter Technical Bulletin 144.

Custom Builders of Electronic Instruments Since 1943



INDUSTRIAL ELECTRONICS, INCORPORATED

Wheeler St., Detroit 10, Mich.



"Fourteen little superintendents..."

Fourteen little angels with eyes glued to the home movie screen. A well planned birthday party proceeding according to schedule. Then ... pfft ... and the projector stops. At once fourteen little angels change to fourteen little devils. Pity the poor host, a mechanical dud, trying to figure out what's wrong. Add fourteen little superintendents and a few assorted mothers, all anxious to take an active part—the perfect set-up for male frustration. When the repairman gets around-after a few days-and the trouble turns out to be electrical insulation failure ... there goes a projector manufacturer's reputation, for a few pennies saved by using an inferior insulation.

Electrical insulation is a decisive factor in successful product operation.

Add extra protection and longer life to your products with BH Extra Flexible Fiberglas* Sleeving (Ex-Flex to the industry.)

Ex-Flex Fiberglas Sleeving has unusual non-fraying properties, obtained independent of the saturant. It can be cut in short lengths, yet will not fray in handling. Its tubular shape makes it easy to handle. It spreads readily over knobs, terminals and irregular objects.

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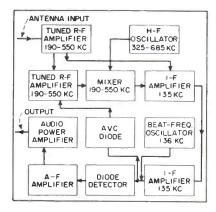


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TUBES AT WORK

(continued from page 138)



Block diagram of the receiver. Compactness is achieved through use of subminiature components

they are double-tuned and use permeability-tuned inductors of about 2.8 millihenries which have Q's of 70 at 135 kc. The resonating capacitors are washer-shaped and mounted in the ends of the transformers. The r-f coil structures resemble those of the i-f transformers so that similar parts may be used in both.

Panel controls involved a design problem. The small size of the panel (5¾ by 1¾ inches) and the need for hermetic seals made it undesirable to have more than two external controls. Since one of these was required for tuning, the other had to fill the three distinct functions of gain control, power on-off switch and beat-frequency-oscillator on-off switch. All these latter functions are performed by a special compact switch, designed to be produced entirely by stamping operations.

The specially designed miniature gain control uses a high-temperature adhesive-tape resistor, also developed by NBS. The tape is applied around a small glass cylinder, is inch in length and inch in diameter, on which 120 axial lines of silver paint have been deposited to form commutator segments. A precious-metal brush makes contact with the projecting ends of these silver lines.

Aside from its inherent usefulness as an ultra compact piece of airborne communication equipment, the new range receiver has also served as the focal point for the development of several novel components and fabrication techniques. These components, engineered to



Study these important design plusses:

TIGHT, DEPENDABLE CONTACTS.

Bolted electrical contacts are firm and substantial. There are no springs to relax because of heat. No risk of the tube slipping from its socket due to vibration.

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Since the GL-6044 has no base in the customary sense, the tube's operating temperature is not limited by decomposition of a basing compound.

BUILT-IN STURDINESS. Strength marks every detail. The tube is short and compact. Its internal structure is solidly braced. Thimble-type terminal seals, of fernicoto-glass construction, are long-

lived, resist strain. Contact fingers, extending out from the tube, have a factor of resilience which helps to ward off shocks.

CLEAN EQUIPMENT LAYOUT. GL-

6044 installation involves no cumbersome conventional socket. No provision need be made for a top anode terminal connection, since all leads come out at the bottom of the tube.

space-saving.GL-6044's are only half the height of 5528 thyratrons—yet rated the same! They will mount more closely side-by-side; their bolted terminals can be interarranged on a panel so that the tubes are compactly grouped.



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Reducing Transmitter Out Time

By James H. Greenwood Station WCAE, Inc. Pittsburgh, Pa.

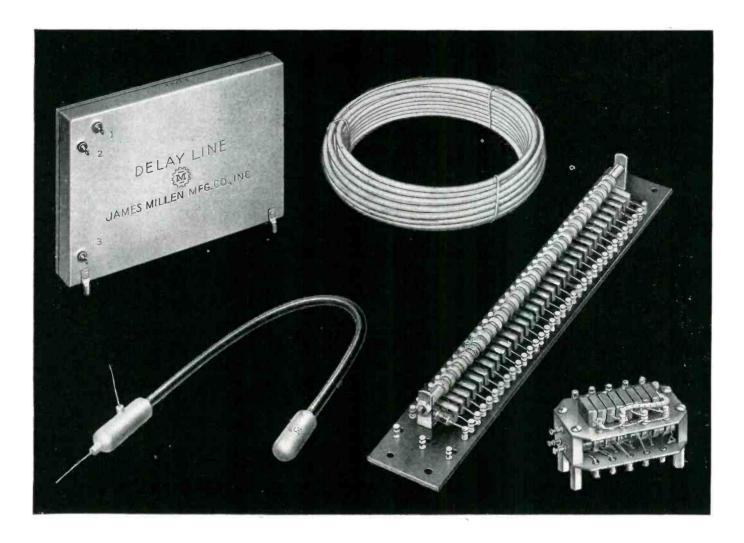
CONTINUITY of service is important to a radio station and the listening public. A minute of silence in the middle of a program may destroy the continuity and annoy the listeners, many of whom would tune in another station.

All of the facilities at WCAE are installed at least in duplicate. Lost time therefore becomes only that time required to isolate the unit in trouble and switch to the alternate unit. This time may be subdivided into four categories; time in which the existence of trouble is observed and realized, time in which to classify or locate the trouble and determine its seriousness, warm-up time for the alternate equipment and actual time involved in switching the alternate equipment into service.

The distribution of time among these classifications varies widely for different types of failure. For some components, the diagnosis and replacement are completely automatic; for others, the human element enters into each step of the process.

At one time, the emergency transmitter at WCAE consumed 45 seconds in warming up. Switching it into service could be accomplished during this time. However, the human element entered into the first two categories mentioned previously and another 15 seconds was added, bringing the total time lost on each occasion of transmitter failure to approximately one minute.

The one-minute loss time has now been reduced to between 12 and 15 seconds. On the surface, it appears to be a gain of four to one but in listener reaction the gain is



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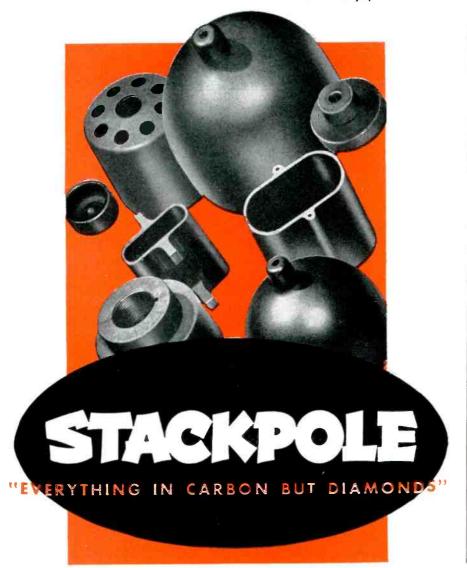


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Because of their durability under heat extremes plus their low secondary emission, Stackpole Graphite Anodes find widespread use in rectifier tubes and, in many cases, surpass hard-to-get molybdenum, tungsten and tantalum anodes in power tube applications. High heat conductivity prevents hot spots with resultant warping. Moreover, graphite permits higher plate power dissipation and safely withstands severe overloads.

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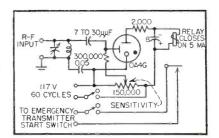


FIG. 1—Schematic diagram of the circuit for starting the emergency transmitter automatically whenever the carrier is interrupted

much greater, for the first quarter minute of silence is not nearly as annoying as the fourth.

Reduced Warm-Up Time

Warm-up time was reduced to 12 seconds by applying heater current continously to all heater-type tubes. This practice has not resulted in any observable reduction in useful tube life since only one heater-type tube has been replaced for any cause since 1939 when the system was installed in its original form.

The human element was removed from observing and realizing the existence of trouble by additional circuitry, as shown in Fig. 1. A small sampling antenna and tank circuit excite the grid of an OA4G. In its plate circuit is a relay which starts the emergency transmitter warm up whenever the r-f grid excitation falls below a safe, preset value. This relay does not apply plate voltage to the emergency transmitter.

If the loss of carrier is caused only by a momentary flashover within a tube or a transient arc caused by a lightning discharge, there is no need for the emergency transmitter and the operator merely shuts it off instead of completing the transfer. Observance of the nature and seriousness of the trouble is accomplished while the equipment is warming up and by the time the decision has been made, the equipment is ready for use.

To reduce the time necessary to switch the alternate equipment into operation, some of the operations performed had to be simplified. The original switching operation consisted of the following steps: removing final-stage plate voltage from the transmitter being taken

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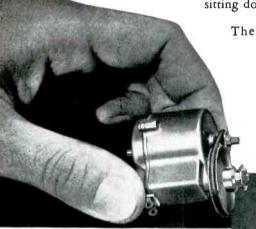
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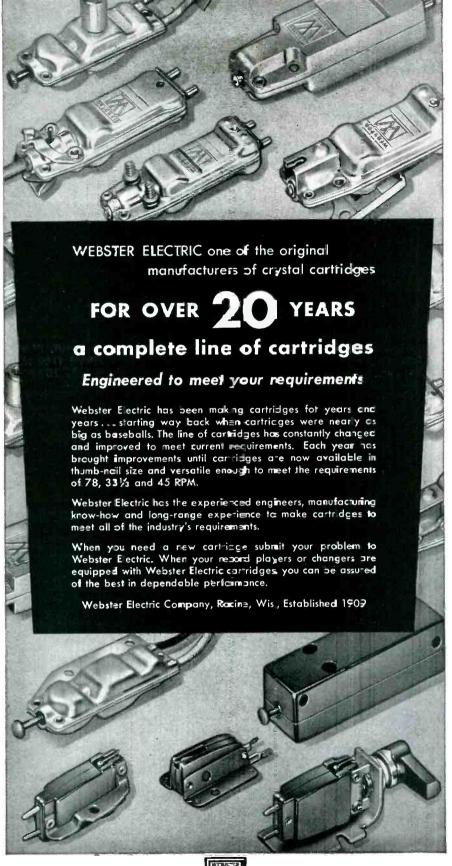
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out of service, switching six circuits from one transmitter to the other (antenna, program input, frequency monitor, modulation monitor, audio monitor and cathode-ray monitor) and applying plate voltage to the transmitter being placed in service.

Controls for most of these operations were grouped at one location. The exceptions were the frequency, modulation and cathode-ray monitors. They were fed from the same sampling antenna which excites the OA4G and therefore indicate operating conditions automatically for whichever transmitter is in use.

With a little practice, all remaining circuits can be switched in a fraction of a second. A completely automatic change-over could have been devised but it would have been considerably more complex and less reliable. Also, it would still require the minimum warm-up time of 12 seconds. The system described retains all of the advantages of human judgment with no added delay.

Pulse-Sharpening Circuit

IT IS OFTEN necessary to sharpen the output pulse of a detector in many applications. Sharpening of the pulse is usually done either to keep a fast discriminator or pulsestandardizing circuit from functioning improperly or to define the time of occurrence of an event more positively.

A pulse-sharpening circuit described by Leonard Reiffel on page 214 of *The Review of Scientific Instruments* for March 1951 was originally designed for use with a Geiger tube but may be adapted for other applications. Figure 1 shows this circuit.

The resonant circuit consisting of the parallel combination of L

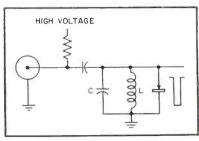
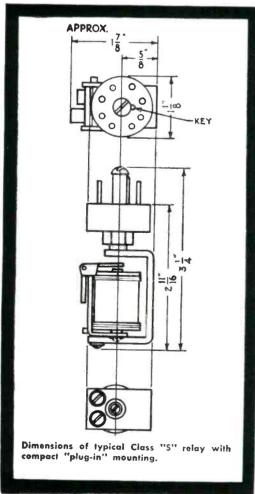


FIG. 1-Pulse-sharpening circuit





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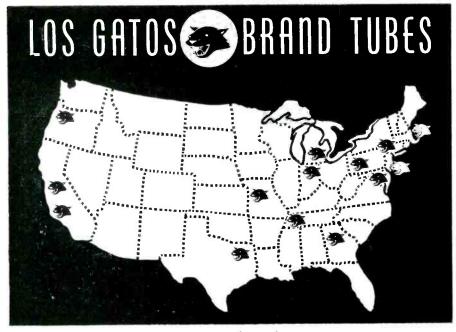
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and C is excited by the current pulse from the Geiger tube. When the resonant circuit is excited, a series of damped oscillations occur. The first half cycle of oscillation is negative and damping is comparatively slight during this period. When the second half cycle of oscillation begins and the polarity is reversed, the germanium diode begins to conduct and heavy damping is placed on the circuit. This damping may be large enough to dissipate all of the energy in the circuit during the first positive half cycle.

If the damping were heavy enough to dissipate the energy during the first half cycle, the output waveform would consist of one-half cycle of oscillation with negative polarity. The output pulse will be very short if the resonant frequency of the circuit is made high.

Constant-Amplitude Oscillator

BY NORRIS C. HEKIMIAN

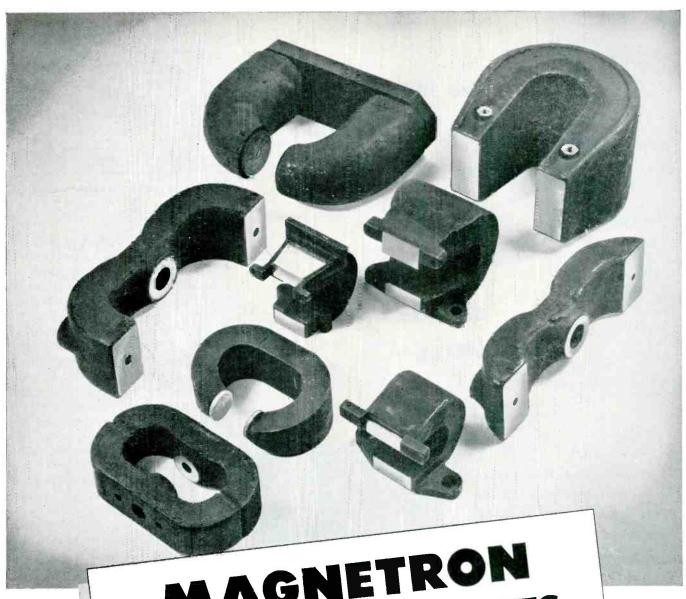
Radio Engineer Central Radio Propagation Laboratory National Bureau of Standards Washington, D. C.

For many applications it is desirable to have a source of r-f voltage which remains reasonably constant in spite of changes in tube parameters, supply voltage, heater voltage and load impedances. Adjustable-frequency oscillators should be stabilized against changes arising from variations in component parts and circuit reactances.

In the past, several systems have been devised to stabilize oscillators, including a modified avc system and fixed-level clipper circuits. The avc system is ineffective in stabilizing class-C oscillators because of the self-biasing feature of such oscil-

The fixed-level clipper has a strong harmonic content in the output waveform. This effect can be reduced by adding a second resonant circuit after the clipper, resulting in increased output impedance and attendant circuit-tracking difficulties

The oscillator to be described does not suffer from the foregoing



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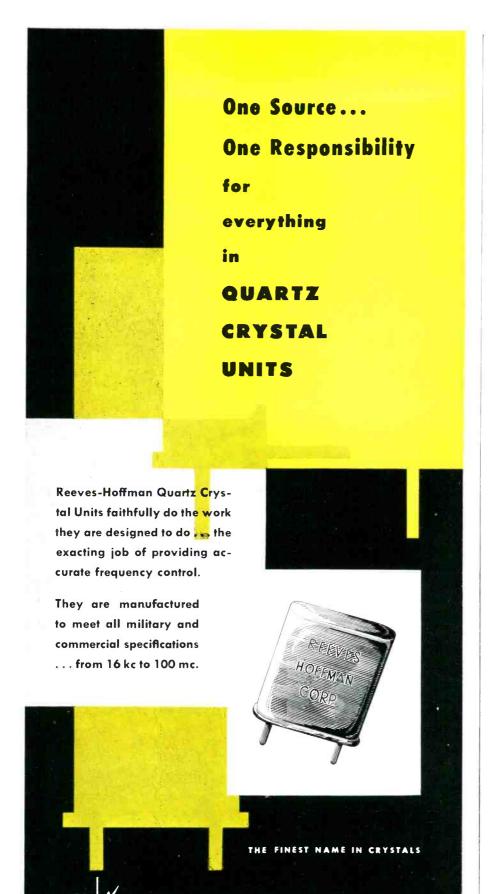
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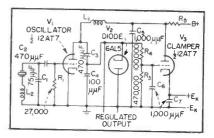


FIG. 1-Clamped oscillator circuit

difficulties to any appreciable extent. It offers the advantage, in application to oscillators which are adjustable over a wide frequency range, of tending to maintain Miller effect capacitance constant by keeping the apparent grid-plate gain of the oscillator reasonably constant. Oscillator frequency calibration drift is minimized by maintaining the tube input and output capacitances practically constant. Clipping of the output waveform is reduced to very low levels. The only clipping that occurs is due to the diode across the output.

The circuit arrangement as shown in Fig. 1, is essentially that of a conventional oscillator followed by a diode across the output terminals. The diode-rectified output is applied to a biased control or clamper tube sharing the same plate-dropping resistor with the oscillator.

With the clamper initially biased in the region of plate cut-off, the rectified oscillator output is applied as a positive voltage to the control tube grid. When the oscillator output voltage reaches a sufficiently high level, the clamper draws plate current and reduces the oscillator plate voltage, thus dropping the voltage to a relatively fixed level.

Normally, the bias voltage E_* is adjusted so that the control tube is always slightly conductive at the minimum level of oscillator voltage anticipated so that positive control action is had at all times. Under these conditions, Fig. 2, the best regulating characteristics are obtained with as large a value of $R_{\rm s}$ as is practical to obtain the desired output voltages.

Figure 3 shows the controlling effect of E_k upon the output voltage. The bias voltage supply for E_k should be of low impedance to

111111111111

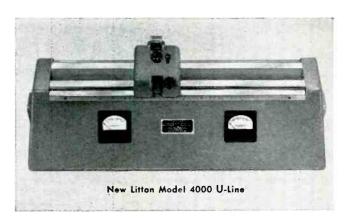
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New Model 4000 U-Line offers utmost convenience and accuracy in quickly determining VSWR in high-power coaxial lines. The equipment transduces power from a standard 15/8" coaxial line to a U-shaped configuration with round central conductor. Both central and outer conductors are mechanically rigid. A traveling probe moves on a precision carriage through the open end of the "U." The probe circuit includes assemblies from the Hewlett-Packard Model 805 Slotted Line. A millimeter scale with vernier indicates probe position. 50 centimeters of travel is available.

HIGH POWER RATING

Model 4000 offers continuous frequency coverage from 450 to 2,750 mc. with insertion VSWR of less than 1.05. Special Teflon bead supports make possible a conservative CW power rating of two kilowatts through the line. VSWR measurements may be made at any power level from kilowatts to microwatts. Standard equipment includes UG-50/U female couplings.

Auxiliary equipment available includes male couplings, VSWR meter, rf power output meter, and range switches to specification.



THERMOPILE

Model 3900 Thermopile is a sensitive, accurate indicator of small-differential temperatures. The unit is equipped with 30 pairs of copper-constantan junctions, and is tapped at 10 and 20 pairs. Uniflare ½" fittings are provided for water connection. Recommended auxiliary meter has a 7-millivolt movement.

ELECTROMAGNET

Model 4807 Electromagnet is a low-voltage, high-current unit designed for general, across-the-line rectified service. It is shell-type in design, and will produce a field of 9,000 gauss across a ½" gap. Model 4807 is the standard Electro-



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WATER LOAD

As a termination for 15/8", 50-ohm coaxial lines, Litton Model 4100 Water Load is particularly useful in highpower applications where power output must be accurately measured. The Load is conservatively rated at 2 kilowatts capacity between 950 and 3,000 mc. VSWR is less than 1.2 over full range, and less than 1.1 above 2,000 mc.

For convenience in sampling rf power, the Water Load also includes two adjustable-depth probes.

For accurate power measurement Litton Model 3900 Thermopile and associated meter are recommended.

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maintain E_* constant. In the test model of the constant-amplitude oscillator, E_* was obtained from batteries with internal resistances of about 5 ohms.

In Fig. 1, C_1 is a frequency-shifting trimmer and L_2 is a crystal-peaking coil employed when using a fifth overtone crystal (30 mc), the particular application for which the constant-amplitude oscillator was developed. Coil $L_{\rm r}$ resonates at the crystal frequency with the circuit capacities shown.

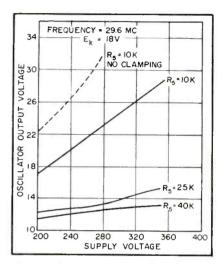


FIG. 2—Oscillator characteristic curves for different values of voltage-dropping resistance

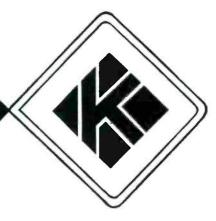
Figures 2 and 3 illustrate the output amplitude stability obtained when the supply voltage, the bias voltage on V_3 and plate-resistor R_5 are varied. Constancy of amplitude versus supply voltage is easily measured and reflects to some extent the increased stabilization against tube parameter changes.

While it is possible to analyze graphically the constant-amplitude oscillator circuit, it is necessary to obtain at least one oscillator characteristic to do so. In most cases it may be advisable to determine circuit values roughly and then adjust the actual oscillator for optimum operation. In practice the output voltage is determined largely by the bias voltage E_k , as shown in Fig. 3, with R_s adjusted for satisfactory regulation. Care should be taken to allow for sufficient dissipation in R₅. A power-type resistor is generally necessary.

Improved clamping may be ob-

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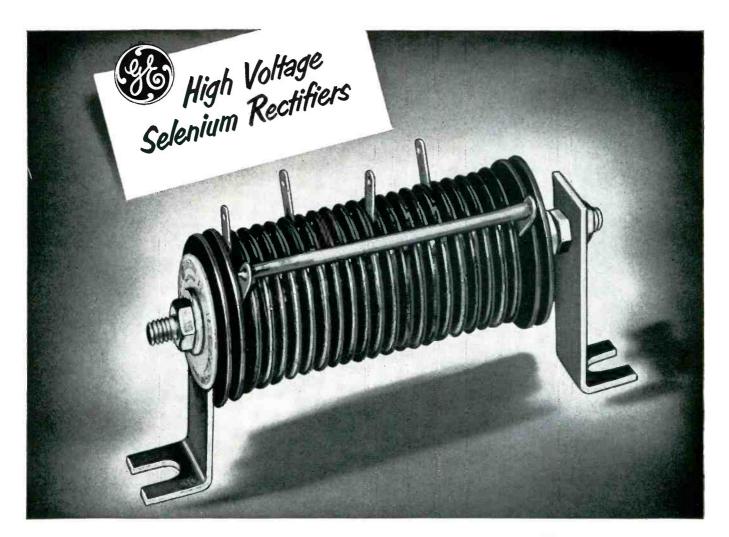
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461-15



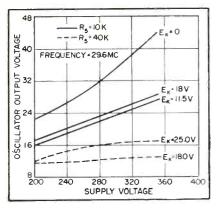


FIG. 2—Oscillator characteristic curves for different values of voltage-dropping resistance

tained by certain circuit modifications. By employing a voltage-multiplier type of rectifier to drive the clamper tube, a greater ratio of d-c control bias to r-f output voltage is obtained with resulting improved regulation. By using a power amplifier as a clamper instead of a type similar to the oscillator, the greater plate-current capabilities result in more positive control action.

If the clipper diode is fed from the final output of oscillator-buffer circuits with more gain included between the oscillator and the rectifier-control circuit, the sensitivity to small changes in output is increased and increased stability results.

Although the circuit described was designed for use as a fixed-frequency local oscillator in a gainstable receiver, it is of value in other applications. Some suggested uses are in exciters, signal generators and high-quality communications receivers.

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fall when it is desirable to have a small amount of heat in the early morning and late evening with the cooling equipment operating during the day. This gives rise to the problem of a suitable control system which will provide just enough heating or cooling to balance the heat loss or gain of the structure, as well as one which can be easily shifted from the heating cycle to the cooling cycle.

A control system of this type must be inexpensive if it is to have a wide application in the field. It is reasonable for a contractor to spend five percent of the contract price on heating and cooling controls, but he could not be expected to spend 15 or 20 percent of the price on a good control system.

The use of electronic controls in the field of heating comfort gave results never before believed practical and this same control system applied to cooling was equally successful. However, the cost of most electronic control systems for year-round heating and cooling application was too large a percent of the contract price on the small business and residential applications for it to have a wide usage.

A modification of the Minneapolis-Honeywell R-7012A Relay for electronic control of home heating has been developed by the author. The modification is inexpensive and gives results closely approximating the ultimate goal for heating and cooling comfort. The original electronic system for heating was described in an article by J. M. Wilson in the December 1950 issue of

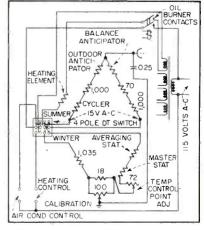
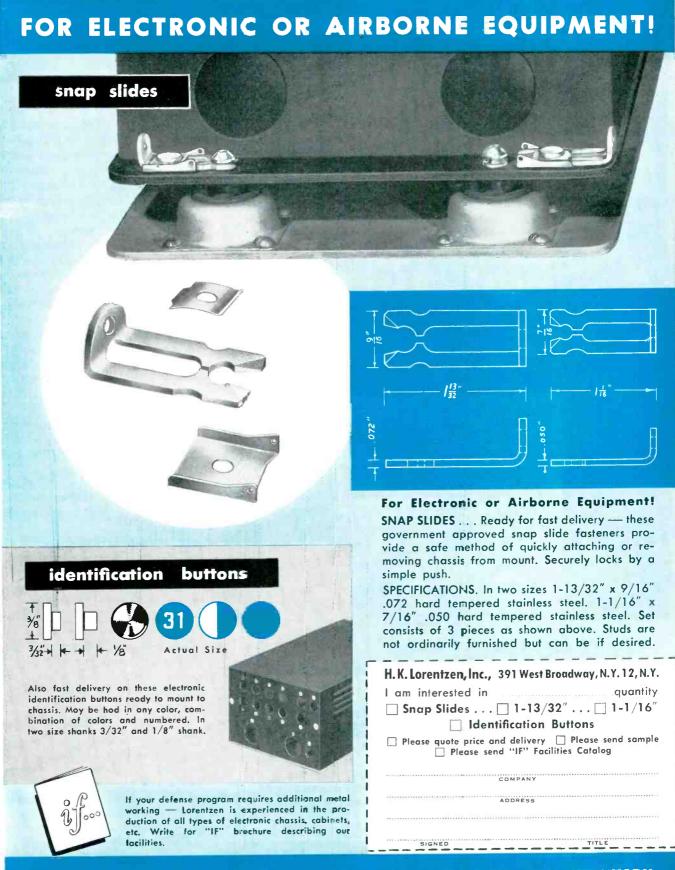


FIG. 1—Circuit modification to Minneapolis-Honeywell R7012A relay for summer-winter application

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ELECTRONICS, page 84. A schematic diagram of the modification is shown in Fig. 1.

Circuit Analysis

Basically, the circuit consists of a Wheatstone Bridge, an amplifier and a relay, with one leg of the bridge sensitive to temperature changes. The lower right-hand leg of the bridge has two resistors in series, a master thermostat and an averaging thermostat. It is intended that the two thermostats be located in different portions of the conditioned area to give an average temperature indication.

The thermostats themselves are temperature-sensitive resistors. When the temperature decreases, the resistors change in value and cause an unbalance in the bridge circuit. A signal voltage is then applied to the grid of the voltage amplifier. When this voltage indicates sufficient unbalance in the bridge, it will cause the relay to close an external circuit controlling the heating device.

As the bridge circuit approaches balance because of the increase of temperature in the controlled area, the voltage across the relay will decrease and the relay armature will drop out, breaking the external control circuit and stopping the furnace.

If the circuit position of the lower right-hand leg of the bridge were exchanged with that of the lower left-hand leg, the reaction previously described would occur in reverse. When the temperature decreased, the signal would decrease and cause the relay to open. Conversely, upon temperature rise a signal increasing in value would cause the relay to close.

Circuit Modification

To make this exchange in position of the legs of the bridge, a four-pole double-throw switch was inserted in the circuit as shown in Fig. 1. It is necessary to connect the relay to the cooling system when the switch is inserted and another pole of the switch is utilized for this purpose.

It was found undesirable to have the cycler operate the relay during the cooling cycle as the recovery of

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new Microline instrument, Model 555 Klystron Signal Source, is an extremely well-regulated power supply. It features a continuously adjustable beam supply from 250 to 3600 volts. In addition, a reflector power supply is continuously variable from 0 to 1000 volts, and a control electrode supply is continuously variable from 0 to 300 volts. The versatility of this signal source permits operation of low voltage as well as high voltage klystrons.

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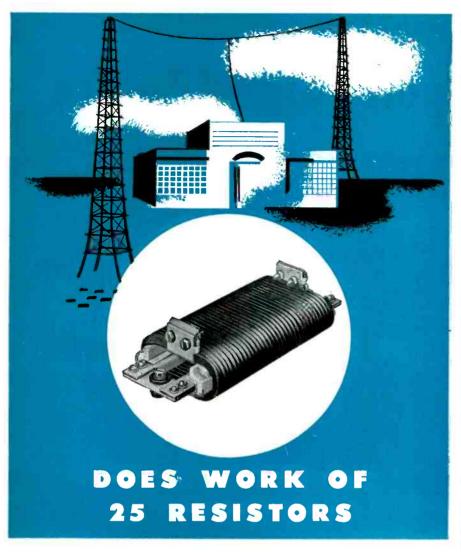
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2K25	3K27	QK-289
2K26	707B	QK-290
2K28	723A/B	QK-291
2K29	726A,B,C	QK-292
2K33	QK-140	QK-293
2K39	QK-141	QK-294
2K41	QK-142	QK-295
2K42	QK-143	QK-306
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saves work—and cost— of hooking them up

It used to take 25 conventional resistors, 11\% x 1\% in., spaced on 2\% in. centers, to keep the power company happy.

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the cooling system is not as fast as that of the heating system. In addition, the human body is not too susceptible to slight variations in temperature during the cooling cycle. To remove the operation of the cycler during cooling, only the circuit to the heater of the cycler was opened by the fourth contact of the switch.

The modified system described, fulfills all requirements in that the modification is inexpensive, it eliminates the usual assortment of heating and cooling thermostats and it makes only one installation necessary.

The system gives more accurate control during the cooling cycle than is possible to obtain with the usual line-voltage thermostat which has a wide differential, usually two to five degrees. The relay when modified retains the desirable characteristics of the electronic heating control during the heating season and when applied to cooling gives a simple one switch change-over.

Thermal Shunt for Soldering Resistors

SOLDERING of resistors in miniature and subminiature equipment is difficult because the resistors may easily become overheated and a permanent change in their values may result. One method developed to help overcome this difficulty is the use of a thermal shunt placed between the body of the resistor and the soldering lug during the soldering process.

A simple way of applying a thermal shunt is to use a pair of long-nose pliers, however, the pliers must not be removed for at least 15

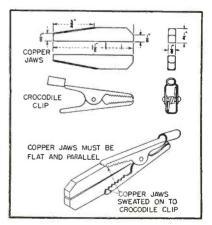
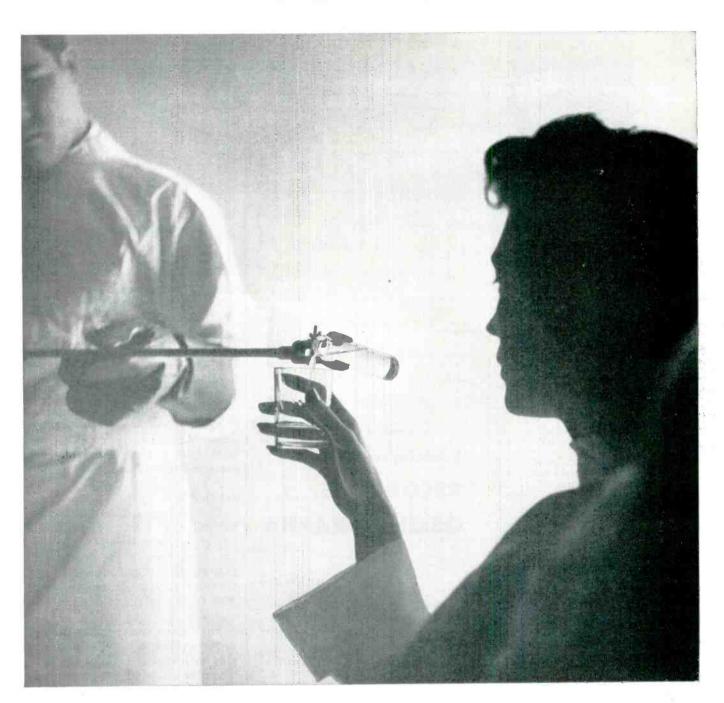


FIG. 1—Component parts and details of the clip-on thermal shunt



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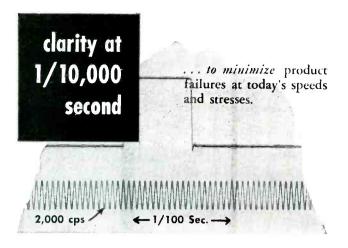
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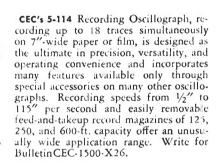
A PERMANENT RECORD is vital in present-day testing, for the millisecond and microsecond have become significant time periods, and interpretation of a phenomenon measured in these units must be accomplished after the phenomenon itself has passed. Consolidated Recording Oscillographs, with record speeds of inches-per-second and stretch the time coordinate to enable the test engineer to read with ease such events and to measure them precisely for evaluation. Consolidated Galvanonieters, with linear frequency response up to 2000 cycles-per-second and more, are employed in the oscillographs to take full advantage of the high record speeds and to yield records on which 0.0002 second may be read with the unaided eve.

The portion of oscillograph record shown was travelling at 100 inches-persecond, each of the vertical timing lines representing 1/100th of a second "stopped" for leisurely, searching analy-

sis. The traces pictured are merely laboratory phenomena. They could as well be the high frequency vibrations and transient stresses which are so important a consideration in today's engineering designs. The test engineer's problem of isolating, measuring and interpreting these phenomena can be answered by Consolidated Recording Oscillographs and Galvanometers. Models are available to fit virtually every precision instrumentation program.

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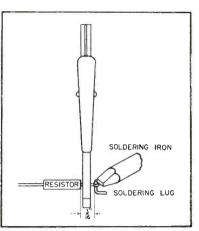


FIG. 2-Sketch showing how the shunt is used when soldering

seconds after the soldering joint has been completed. Copper pliers are more effective than steel ones because of the better conductivity but, in either case, the operator would tend to remove the pliers at the same time the soldering iron is removed. This practice allows the heat stored in the molten solder, the terminating lead and in the soldering lug to flow freely into the resistor body.

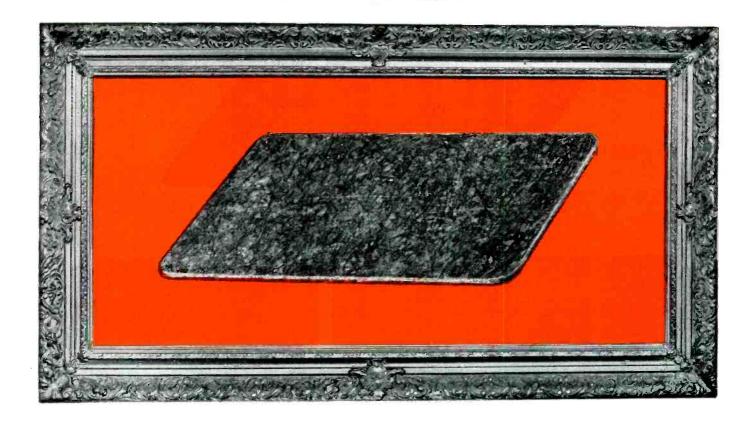
To guard against the operator removing the shunt too quickly, a clip-on device is desirable. The clipon action compels the operator to remove the shunt forcibly and thereby tends to prevent the operator from removing the shunt prematurely.

The clip-on device finally decided upon utilizes a simple crocodile clip with copper bars in its jaws. The dimensions of the copper bars are 1-in. thick, 2-in. wide and 12-in. The component parts and details of the clip-on thermal shunt are shown in Fig. 1. Figure 2 shows a sketch of the shunt in use.

The information in this article was abstracted from a report by E. N. Shaw entitled "The Overheating of Miniature Resistors During Soldering" published by the Telecommunications Research Establishment of Worcester, England.

Position Indicator

A RECENT DEVELOPMENT in the field of position indication consists of magnetic measuring system coupled to an electronic circuit for control or remote indication purposes. The instrument is espe-



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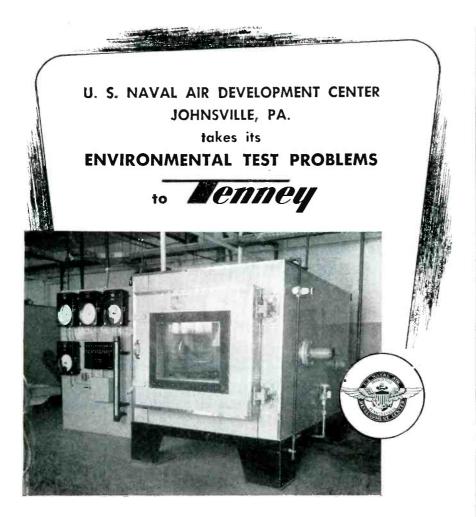
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cially suited for the indication of liquid levels particularly under high pressures and where seals would be undesirable.

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The magnetized strip is pivoted horizontally at a point in its length which corresponds to the normal level of liquid in the chamber with a suitable adjustment for height. The strip is mounted so that its magnetic poles and those on the floating magnet repel each other. The floating cylinder therefore rolls against the guides without friction. When the cylinder leaves its normal position, it tends to repel the strip magnet and cause it to rotate slightly one way or the other and close a pair of contacts fixed to it.

The contacts are connected to the grid circuit of an amplifying tube whose output may be used to operate automatic control equipment to change the liquid level or operate remote indication devices.

A restoring force for the strip magnet and contact arm is provided by electromagnetic or pneumatic means.

Measuring Muzzle Velocity

CONVENTIONAL equipment used for testing muzzle velocities of large-calibre pistols utilize magnetic pickups. Such equipment is not applicable for determining the muzzle velocity of pistols firing pellets of the order of 0.177 calibre.

The problem of determining the muzzle velocities of air pistols firing small-calibre lead pellets may be solved by the arrangement shown in Fig. 1. Each photocell is located opposite a narrow-beam light source. The air pistol is mounted in a cradle and aimed so that the pellet passes between the photocells and the light sources.

When the pellet passes between

THIN ELECTRICAL STEELS FOR OPERATION at 400 Cycles

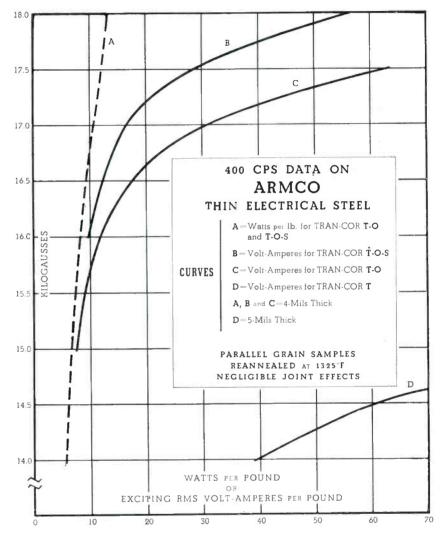
Two new thin (4-mil) oriented electrical steels for operation at 400 cycles have been developed by Armco Research. They are known as Armco TRAN-COR T-O and TRAN-COR T-O-S.

TRAN-COR T-O was developed for use at 400 cps and higher with inductions up to 16 kilogausses. TRAN-COR T-O-S is recommended for equipment where more perfect orientation permits operations at still higher induction without excessive excitation.

The new TRAN-COR T-O-S can be operated at inductions 20% higher than any nickel-iron alloy. It was developed under a contract with the Bureau of Aeronautics of the Navy Department to make possible a great reduction in weight of airborne electrical equipment.

Curves on this page show a comparison of the excitation required by the two grades of oriented 4-mil material. TRAN-COR T-O-S is so carefully controlled in processing that the minimum permeability at 18 kilogausses is 1800. This value is not usually attained even at 14 kilogausses by the best grades of non-oriented high silicon steel.

TRAN-COR T-O-S has a density of 7.65, a volume resistivity of 47 microhm cm. or 282 ohms/mil ft., and a lamination factor of 95% solid.

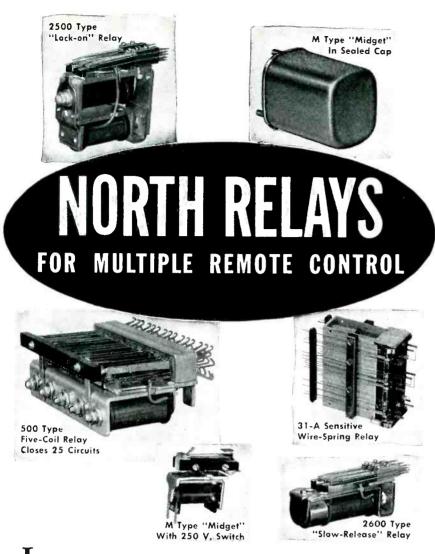


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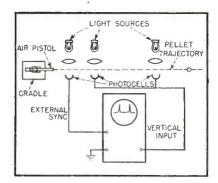


FIG. 1—Sketch of the arrangement for measuring the muzzle velocity of air pistols

the first photocell and light source, interrupting the beam of light, the low-speed driven sweep of the oscilloscope is initiated. As the pellet passes between the second and third beams, the electrical impulses generated are applied to the vertical deflection plates of the oscilloscope. These impulses appear on the screen as vertical deflections.

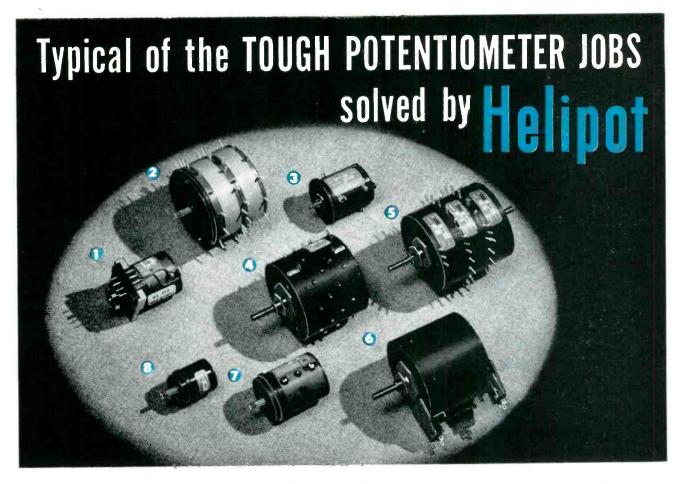
The muzzle velocity can be computed by measuring the distance between the two vertical deflections. For continual use in a production setup, the limits representing maximum and minimum tolerable muzzle velocity may be marked on the screen. By so doing, testing can be done by wholly unskilled workers.

This material was abstracted from the 1950 July-September issue of *The Oscillographer*, published by the Allen B. DuMont Laboratories, Inc.

Robot Control for TV Transmitter

AN EXPERIMENTAL robot control for the new BBC tv transmitter at Daventry, England, has been incorporated in the control room. The robot rings an alarm bell if anything goes wrong with the transmitter and engineers in the transmitting building several hundred yards away or in London can switch over transmitters by means of a system of telephone signals.

A giant fan and air-cooled valves are used in the new transmitter. The fan brings in fresh air and uses the heated air, if necessary, to circulate through the station for heating purposes.



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The potentiometers illustrated abave are typical examples of the tough problems HELIPOT engineers are solving every day for modern electronic applications. If you have a problem calling for utmost precision in the design, construction and aperation of potentiometer unitscoupled with minimum space requirements and maximum adaptability to installation and operating limitations—bring your problems to HELIPOT. Here you will find advanced "knaw-how," coupled with manufacturing facilities unequaled in the industry!

The HELIPOTS above—now in production for various military and industrial applications—include the fallowing unique features . . .

This 10-turn HELIPOT combines highest electrical accuracies with extremes in mechanical precision. It features zero electrical and mechanical blacklash...a precision-supported shaft running on ball bearings at each end of the housing for low torque and long life ... materials selected for greatest possible stability under aging and temperature extremes ... special mounting and coupling for "plug-in" convenience ... mechanical and electrical rotation held to a tolerance of $\frac{1}{2}$ ° ... resistance and linearity accuracies, ±1% and ±0.025%, or better, respectively.

This four-gang assembly of Model F single-turn potentiometers has a special machined aluminum front end for servo-type panel mounting, with shaft supported by precision ball bearings and having a splined and threaded front extension. Each of the four resistance elements contains 10 equi-spaced tap connections with terminals, and all parts are machined for greatest possible stability and accuracy.

This standard Model A, 10-turn HELI-POT has been modified to incorporate ball bearings on the shaft and a special flange (or ring-type) mounting surface in place of the customary threaded bushing. This HELIPOT also contains additional taps and terminals at the 1/4- and 93/4-turn positions.

This standard Model B, 15-turn HELI-POT has a total of 40 special tap connections which are located in accordance with a schedule of positions required by the user to permit external resistance padding which changes the normally-linear resistance vs. rotation curve to one having predetermined non-linear characteristics. All taps are permanently spot-welded and short out only one or two turns on the resistance element-a unique HELIPOT feature!

This six-gang assembly of standard Model F single-turn potentiometers has the customary threaded bushing mountings, and has shaft extensions at each end. The two center potentiometers each have 19 equispaced, spot-welded tap connections brought out to terminals. Each tap shorts only two turns of .009" diameter wire on the resistance element.

15 This Model B, 15 turn HELIPOT has been modified to incorporate, at the extreme ends of mechanical and electrical rotation, switches which control circuits entirely separate from the HELIPOT coil or its slider

This 10-turn HELIPOT has many design features similar to those described for unit No. 1, plus the following additional features . . . a servo-type front end mounting ... splined and threaded shaft extension ... and a center tap on the coil. All components are machined to the highest accuracy, with concentricities and alignments held in some places to a few ten-thousandths of an inch to conform to the precision of the mechanical systems in which this HELIPOT is used. Linearity accuracies frequently run as high as ±0.010%!

This single-turn Model G Potentiometer has been modified to incorporate a ball bearing shaft and a servo-type front end mounting. Special attention is given to contact designs and pressures to insure that starting torque does not exceed 0.2 inch-ounces under all conditions of temperature.

The above precision potentiometers are only typical of the hundreds of specialized designs which have been develaped and produced by HELIPOT to meet rigid customer specifications. For the utmost in accuracy, dependability and adaptability, bring your potentiometer problems to HELIPOT!

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Engineering bulletins describing each of the many basic types of connectors are available. We will gladly send you any of these if you will simply describe your connector requirements.

Molded Polychloreprene inserts 75-80 shore hardness provide pressure-proofing of both pin and socket con-tracts. Have high dielectric strength under wide range of temperatures and at extreme altitudes. Mated fit-tings will not show more than 10 microamperes dielectric leakage and will not arc when subjected to 7500v dc at room temperature.

Pin Contacts machined from solid brass, silver-plated. Solder cup hand-tinned.

Machined ball-in-cone joints provide radio shielding and improve vibration resistance.

Socket contacts machined from solid copper alloy with new Cannon design, silver-plated.

Matching serrations in end bell and shell make practical wrench-tighten-ing from one side of the installation without putting strain on contacts or

Polychloreprene grommets make moisture-proof seal over soldered connections

Concentric rubber bushings under pressure of cable clamp provide snug, moisture-proof wire entry. Eliminate usual strain on outer wires Provision is made for grounding lug

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THE ELECTRON ART

(continued from p 140)

exceed 5,000 volts.

num film and excite the phosphor. In general, excitation occurs only over a short portion of alternate half-cycles. It occurs on the negative swings, because the penetrating power of the electrons is greater than that of the positive ions. At radio frequencies the number of exciting pulses per second is sufficiently large to produce, as far as the eye can detect, a uniformly illuminated screen. The frequency of the exciting pulses is not critical; in apparatus devised by the author, it is of the order of one megacycle. The r-f voltage must

Although the electron and ion current during excitation of the screen is of the order of a milliampere, the displacement current may be 100 times as large. It is important, therefore, that care be exercised to minimize stray capacitance between the probe and ground, especially if several tubes are to be operated simultaneously from the same oscillator.

Small bulbs or enclosures containing a phosphor screen may be successfully examined simply by plugging the stem with a rubber stopper containing two holes, one in the center for a metal probe, the other for a piece of glass exhaust tubing. After the pressure range

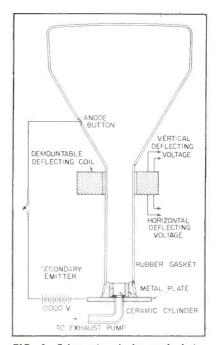


FIG. 2-Schematic of d-c method for testing picture tube screens

40,000,000 watts

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...with the help of LOW CARBON NICKEL

Specifications for the Armed Services VC-1257, a large hydrogen thyratron tube, call for an output of 40 megawatts. This requires tube operation at 40,000 volts, at a peak current of 2000 amperes.

When approached with the problem, engineers at Chatham Electronics Corp., Newark 2, N. J., knew that the metal parts of the tube should have the following qualities:

- 1. A low carbon content to prevent the hydrogen in the tube from forming hydro-carbons.
- 2. Good de-gassing properties.
- 3. High resistance to corrosion.
- 4. Good workability and weldability.

At the suggestion of INCO's sales and engineering men, Chatham

tried Low Carbon Nickel (carbon content 0.01 to 0.02 per cent), for the tube parts where good workability was one of the important features. The metal proved more than satisfactory, and full advantage of its favorable characteristics was taken in designing certain critical tube parts. Today the Armed Services' VC-1257 tube has one of the largest peak power outputs on record.

Supplying information on suitable alloys that solve problems such as corrosion resistance, heat resistance and high purity is the function of INCO's Engineering Service. Although metals are hard to get at the present time, our engineers will be happy to consult with you on your particular design problems, if you will write, stating your requirements.



Low Carbon Nickel is produced under controls assuring no more than a 0.02% carbon content. It is resistant to corrosion and can be used in oxidizing temperatures up to 2000°F. Softer than pure nickel, it is slow to work harden. It is chiefly used for difficult deep-drawing and spinning operations.

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LOW C		NICKEL (ANNEALED) PROPERTIES	
Yield Strength (0.2% offset) 1000 psi	15	Specific Heat (32°-212°F.) Btv./lb./°F.	0.11
Tensile Strength 1000 psi	40	Thermal Expansion Coefficient (32°-212°F.) 10 ⁸ in./in./°F.	7.2
Elongation in 2 in., per cent	50	Thermal Conductivity (32°-212°F.) Btu./sq. ft./hr./°F./in.	420
Hardness Brinell	90	Electrical Resistivity (32°F.) ohms/cir. mil. ft.	50
Density 1b./cu. in.	0.321	Tensile Modulus of Elasticity 10— ⁶ psi	30
Melting Point (°F)	2615-2635	Torsional Modulus of Elasticity 10 ⁻⁶ psi	11

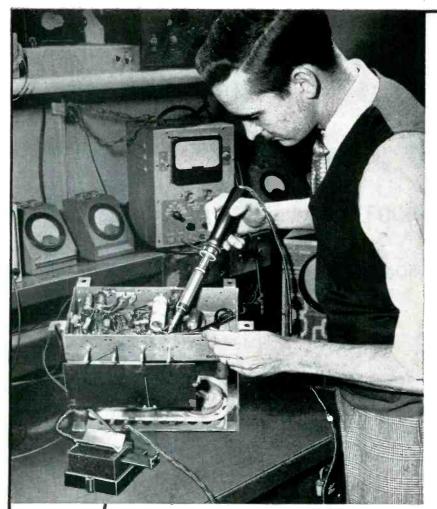
LOW CARBON NICKEL

Limiting Composition %

N	Co)	99.00 min.
Cu		.25 max.
Fe		.40 max.
M		.35 max.
C		.02 max.
Si		.35 max.
S		.01 max.



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has been reached, an ordinary spark coil may be touched to the external lead of the probe. For large bulbs more power is needed for stable operation, and it is advisable to use some type of radio-frequency power oscillator having an output of 100 to 200 watts.

For tests on picture tubes having metal shells, the r-f oscillator and transformer can be replaced by a high-voltage, low-current, 60-cycle transformer. In this test, the metal shell is grounded. This circuit arrangement is not satisfactory for tubes having glass bulbs, because the high current through the anode button would probably damage the connection between the conductive coating and the button.

Direct-Current Method

In the d-c method (see Fig. 2) the vacuum seal of the bulb is made in the same way as that described for the r-f method, except that a ceramic cylinder rather than a metal one must be used to support the gasket in order to prevent arc discharges during the test. The ceramic and rubber gasket are held in place by screws fastened into the ceramic from the metal plate and then soldered. The surface of the metal plate inside the ceramic cylinder is covered with a flat disc composed of some type of secondary emitter such as sodium silicate with just enough micronized graphite added to give the silicate some conductivity.

To carry out the test, the operator opens the valve to the exhaust pump and exhausts the bulb to a pressure within the range of 5 to 20 microns. When this pressure range has been reached, a positive voltage of approximately 10,000 volts with respect to the metal plate is impressed on the anode button of the bulb. Within this pressure range sufficient ionization occurs to maintain the desired excitation. At higher pressures the system has a tendency to arc; at lower pressures the excitation becomes too faint or ceases. In normal operation positive ions produced in the bulb by collisions strike the secondary emitter and liberate electrons. These electrons form a beam which. collimated by the bulb neck, pro-

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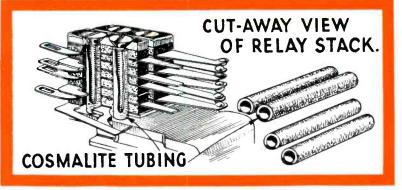


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duces an illuminated spot on the phosphor screen just slightly larger in diameter than the diameter of the bulb neck. This beam may be deflected by a demountable scan-The deflection coils ning yoke. may be energized by sinusoidal voltages, provided that the voltages are sufficient to considerably overscan. Under this condition, the scanning rate over the faceplate is relatively constant, so that the illumination is nearly uniform. One voltage can be obtained from the regular 60cycle line and the other from an audio power oscillator operating at 12 to 15 kilocycles. The use of sinusoidal waveforms instead of sawtooth waves simplifies the circuitry without sacrificing any features needed for detecting flaws in the screens.

Current Transformers for Audio Measurements

By Lawrence T. Fleming National Bureau of Standards Washington, D. C.

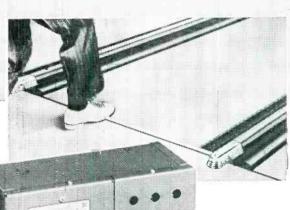
A WELL KNOWN METHOD for the measurement of heavy current at low frequency involves the use of a current transformer with a rectifier ammeter connected to its secondary winding. The feasibility of using this method for measuring a current of the order of a few amperes is probably frequently overlooked. Moreover, measurements made on an improvised arrangement of this type have shown fair accuracy over the range 50 to 50,000 cycles per second.

This note is written to call attention to the following attractive features of this method: (1) the scale is nearly linear; (2) the instrument can withstand considerable overloading without injury; (3) the power required for indication is less than for thermal-type ammeters. One may improvise an arrangement of components on hand in many laboratories that will give satisfactory performance over the audio range.

Thus for the measurement of the order of one to three amperes we have improvised a current transformer through the use of a Seletron SELENIUM RECTIFIERS

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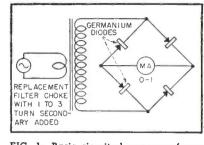


FIG. 1—Basic circuit shows use of current transformer for audio-frequency measurements

replacement choke, constructed a suitable bridge-type rectifier using germanium diodes, and for current indication used a one-milliampere d-c instrument. Figure 1 shows the circuit arrangement. In one case the rectifier-instrument combination employed four-IN34 crystal diodes and a 1-milliampere 50-ohm indicating instrument.

The transformer was made by winding 1 to 4 turns around the outside of the winding of an ordinary radio replacement-type choke coil and interleaving the core laminations to secure low reluctance and thus low magnetizing current. The secondary of the transformer had an inductance of about 40 henrys after interleaving and a transformer ratio of the order of 1 to 1,000.

Such an arrangement in which the reactance of the secondary of the transformer is quite high, offers the advantage that the d-c milliammeter may be shunted by a resistor to secure the desired nominal value of current ratio for the combination. This procedure has little adverse effect on frequency response of the combination for adjustments as large as 30 percent of the full scale value of the milliammeter. Further adjustment may of course be made by changing the number of primary turns.

It should be noted that the transformer is operated as a true current transformer, because the equivalent impedance of this combination as seen from the primary will ordinarily be low compared to the impedance of the source, and the secondary reactance is high compared to the impedance of the rectifier-instrument load connected across it. In this respect the arrangement is quite like the much



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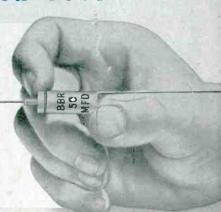
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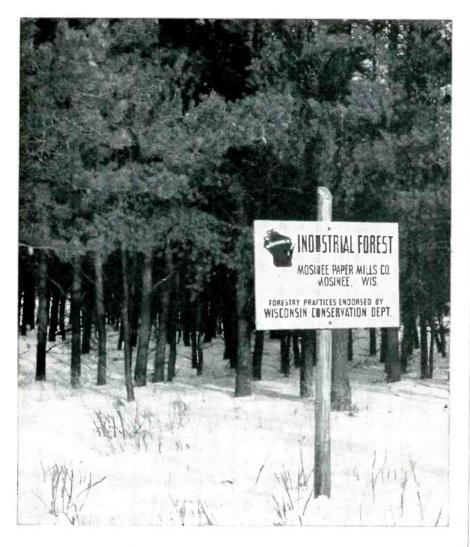


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larger current transformers used in measuring large currents at power frequencies. Its purpose is to provide an impedance and voltage level at the secondary which is suitable for the linear and efficient operation of a rectifier instrument.

The essential requirement in the transformer design is that the secondary reactance be high compared to the total d-c resistance in the secondary circuit, at the lowest frequency at which the instrument is to be used.

High-Frequency Limitations

In Fig. 2 are shown experimental curves of voltage drop and instrument error with respect to frequency, for an arrangement of the kind shown in Fig. 1. The error at low frequencies depends on the ratio of the impedance of the rectifier instrument to the impedance looking into the secondary of the transformer, as will be discussed later. Error does not begin to be noticeable at high frequencies until slightly above 50 kc. An increase in voltage drop across the primary of the transformer is, however, evident some time before this, beginning at 10 kilocycles.

The high-frequency limitations cannot be readily estimated on paper because at high audio frequencies the effect of distributed capacitance in the secondary coil windings will be to shunt secondary current away from the rectifier instrument and thus change the effective current ratio. Essentially the method should be good at the higher frequencies because the rectifier instrument impedance is comparatively low and in effect constitutes a short circuit of the highimpedance secondary winding. Thus the voltage developed across sec-

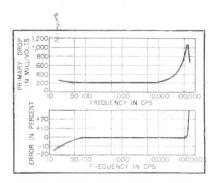


FIG. 2—Curves show error and primary voltage drop at different frequencies



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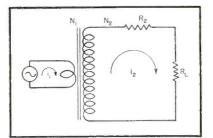


FIG. 3—Equivalent circuit used in calculating low-frequency performance

tions where distributed capacitance is high would normally be quite low. Secondary resonance should therefore only become troublesome at rather high frequencies.

Design improvements should be possible through minimizing secondary leakage reactance, through reducing the number of primary and secondary turns (commensurate with the required secondary reactance) in order to reduce stray and distributed capacitances, and through the proper selection of core material if it were desired to increase the upper frequency limit. On the other hand this might restrict the low-frequency useful limit and make it desirable to have two transformers with overlapping frequency ranges-one for low frequencies and the other for very high audio frequencies for use with the rectifier-type instrument. It is very likely that the frequency range can be substantially widened by employing high permeability core material, and reducing the number of secondary turns to the minimum required for the necessary inductance.

This method has the inherent wave-form limitation of rectifiertype instruments in that the deflection of such instruments are proportional to the average rather than the effective value of current. Thus if calibrated on a sine wave to read effective values they do not indicate properly on distorted wave-forms. The arrangement, however, is not particularly sensitive to change in resistance of the rectifiers with temperature as this effect is swamped by the high reactance of the transformer secondary. Except on distorted wave-forms an arrangement of this sort should be considered in the 5-percent-accuracy class unless special precautions are taken in its calibration.

The circuit behavior for low fre-



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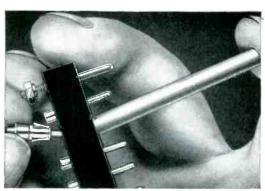
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Hillside 9294

quencies may be analyzed as follows:

Referring to Fig. 3, the current

$$\frac{I_2}{I_1} = K \frac{N_1}{N_2} \left(\frac{1}{1 - j \frac{R_2 + R_L}{\omega L_2}} \right)$$

where I_1 = primary current, I_2 = secondary current, K = coupling coefficient, N_1 = primary turns, N_2 = secondary turns, R_2 = resistance of secondary winding, R_L = secondary load resistance, and L_2 = secondary inductance. To retain a constant ratio we must have

$$\omega L_2 >> (R_2 + R_L)$$

This determines the lower frequency limit. For example, for a 5-percent error at the low frequency end of the range,

$$\left[1 + \left(\frac{R_2 + R_L}{\omega L_2}\right)^2\right]^{\frac{1}{2}} = 1.05$$

Approximately

$$\left(\frac{R_2 + R_L}{\omega L_2}\right)^2 = 0.1$$

SC

$$\omega L_2 = \frac{R_2 + R_L}{0.31} = 3.1 (R_2 + R_L)$$

If $R_2 + R_L = 600$ ohms, and the low-frequency limit is taken at 20 cps, L = 14.8 henrys.

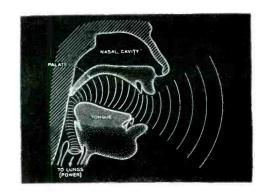
REFERENCE

(1) Edgecombe and Ockenden, "Industrial Electrical Measuring Instruments", p 175, Sir Isaac Pitman and Sons, 1933.

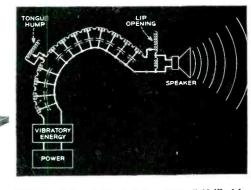
Improved Terratex

CONSIDERABLE ACTIVITY has been devoted to the development of a satisfactory mica substitute. It has been the hope of those working in this field to develop a material having all the physical and chemical characteristics of mica. We are all more or less familiar with the progress which has been made in this field, and certainly that progress has been significant. Nevertheless, an easily producible material satisfactory for vacuum-tube applications, which are very demanding indeed, has not yet been reported.

In a paper presented to the AIEE in 1948, T. R. Walters, of G-E Works Laboratory at Pittsfield, reported the original development of a material known as terratex. The







The machine at the left is saying "Ah!" It's the new electrical vocal system developed at Bell Laboratories. Top sketch shows human vocal system also saying "Ah!" The electrical model is sketched below it. Energy source at bottom of "tract" can emit a buzz sound, like vocal cord tone, or the hiss sound of a whisper.



No one else speaks exactly like you. Each of us uses different tones to say the same words. To study and measure how we make speech, acoustic scientists of Bell Telephone Laboratories built a model of the youal system.

Electric waves copy those of the vocal cords, electric elements sim-

ulate the vocal tract, and, by adjustments, vowels and consonants are produced at pitches imitating a man's or woman's voice.

Using this electrical system, telephone scientists will be able the better to measure the properties of people's voices. Knowing more about speech they can find better and cheaper ways to transmit it.

This is another step in the research at Bell Telephone Laboratories which pioneered the exact knowledge of speech. Past work in the field is important in today's fine telephone service. A still deeper understanding of speech is essential in planning for tomorrow.

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Until recently there has been no one place where components specifically designed for plug-in unit construction were available. It was necessary for engineers to design and have parts custom made or improvise with standard components in makeshift arrangements. To provide the type of design necessary, Alden engineers are working with the industry developing a whole series of components specifically for plug-in

construction.

The first problem undertaken by Alden engineers was a base specifically for plug-in unit construction... The conventional tube type bases proved unsatisfactory; they didn't stand up, the boss broke and the pins bent. To overcome these difficulties Alden designed an entirely new base... the Non-Interchangeable Series bases have no molded center boss to break, pins

are strong and stubby—do not bend or break out and are Non-Interchangeable to prevent danger of mismating and costly burned out units.

Out of this work we feel that Alden's is the one place where you now can take your unitizing problems and obtain the standard bases, sockets, mountings and housings to answer most of your needs. As illustrated below, the Alden Non-Interchangeable and miniature bases have tremendous flexibility and are fast becoming the standard for plug-in construction.





The scope of the Alden "20" base as a mounting medium is almost unlimited . . . cards, brackets and bails can be easily and securely attached with standard assembly tools. For holding components and miniature tube sockets the Alden Terminal Card Mounting System on the Alden Base gives ease of layout and wiring assembly. Open units for heat dissipation or shielded units for protection against dust or rough handling both lend themselves to mounting on the Alden Base with the same facility.

11 Pin Non-Interchangeable Bases & Sockets

Smaller than the "20" but with the same features, the Alden "11" base and sockets are rugged for long life and Non Interchangeable to isolate critical voltages or signals and prevent burned out units. The retention force of pins and socket clips can be varied from light to heavy. Locating rings and alignment indicator quickly center base and socket for insertion. These and other features make it practical to incorporate plug-in construction in your design.

7 and 9 Pin Miniature Bases & Sockets

Miniature and sub-miniature circuits, potted circuits, and miniaturized components easily become compact, sturdy plug-in units with the Alden 7 and 9 pin miniature plug-in bases and sockets. A wide selection of housings and mounting components are available for use with these bases.

Of particular importance is the Alden Tor-

Of particular importance is the Alden Terminal card Mounting System. Miniature circuits can be assembled on the card and the assembly can be mounted on the base to form a complete miniature unit.

Write Dept. E for information on sample plug in kits or for Catalogue "Components for Plug-in Unit Construction."



ALDEN PRODUCTS COMPANY

117 NORTH MAIN STREET, BROCKTON 64, MASSACHUSETTS

untreated paper-like sheet of terratex is mechanically weak, but a process has been developed to give it desirable physical characteristics. To make it suitable for use in vacuum tubes, the spacer material is impregnated with partially hydrolyzed ethyl silicate which, when dried, vields the tough terratex that is usable for spacers. Care must be taken in the choice of hydrolyzing acids used when forming silicatetreated terratex for vacuum tube use so that contaminants are not introduced which would be harmful in the tube. Although the processing of the spacer material is not yet complete at this stage in the manufacturing process, the terratex in this condition can be stored or shipped conveniently without noticeable deterioration.

Vacuum tube spacers have some dimensions with very close tolerances. To assure reasonable uniformity of spacers, the material should punch cleanly and easily and should not flake or split when handled. In addition, it should retain a certain amount of resiliency in order for the points of the spacer which contact the bulb of the miniature tube to yield slightly as the mount is inserted in a bulb which is slightly out of round or undersize. Chemically, the material should be stable so that storage problems are minimized and so that uniformity can be achieved from one tube production lot to another. Electrically, the material should have high resistivity and a low dielectric constant. In addition, for vacuum tube applications considerable surface roughness is desirable

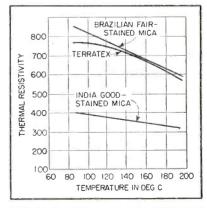


FIG. 1—Thermal resistivity of terratex is almost the same as that of Brazilian fair-stained mica



This new service offers the same "Precision Production" that made Federal the outstanding name in H-F cables

MORE than good cable and good connectors are needed to make good assemblies. The secret of their efficiency and dependability is the skill with which the two are joined!

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An important feature of Federal's new service for users of coaxial cables is the fact that Federal holds approval of the U. S. Signal Corps and the U. S. Air Force for the new low-temperature, non-contaminating thermoplastic jacketed cables—an original development by Federal. For quick action on quotations, send us your detailed drawings now—with full information on requirements. Address your inquiry to Dept. F-513.

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Export Distributors: International Standard Electric Corp., 67 Broad St., N. Y.





in order to minimize the development of leakage paths during the operating life of the tube.

For vacuum tube applications, particular characteristics of the spacer material are of special significance. A comparison of the thermal resistivities of mica and terratex is shown in Fig. 1. From these curves it can be seen that terratex has approximately the same thermal resistivity as Brazilian fairstained mica and is somewhat superior to Indian good-stained mica. The characteristics are, of course, difficult to measure accurately due to the variation in air spaces in the mica specimens. An interesting comparison of performance is that the cathode temperatures are about the same for tubes spaced with 20-mil terratex or 10-mil mica.

The effects of temperature upon the insulation resistance of mica and terratex are essentially the same over the normal temperature range for the mica and terratex specimens. Measurements have been made of power factor and dielectric constant to 50 megacycles. At this frequency the power factor of mica and terratex are about the same, and the dielectric constant is approximately 3.5 as compared to the mica constant of 4.5 for the specimens measured.

Figure 2 shows an operating time comparison between the transconductance of type 12AT7 tubes spaced with micas and the same type tubes, identically processed, spaced with processed and plain terratex spacers. It can be seen from these curves that the results are quite comparable and satisfactory. Those familiar with the characteristics of the type 12AT7 probably realize that this is a tube extremely susceptible to modifica-

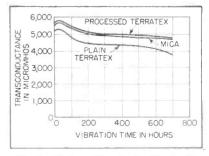


FIG. 2—Curves show desirable characteristics of terratex for use in vacuum tubes that will encounter vibration



BRIDGEPORT BRASS COMPANY

COPPER ALLOY BULLETIN

"Bridgeport

MILLS IN BRIDGEPORT, CONN. AND INDIANAPOLIS, IND. —IN CANADA: NORANDA COPPER AND BRASS LIMITED, MONTREAL



Continuous Casting of Brass at Bridgeport Brass Co., Bridgeport, Conn.

Brass Scrap Situation and Its Effect on Brass Production

Copper and zinc, essential metals for both military and civilian needs, have been in short supply for many months and the situation has been growing worse. After taking care of the mounting volume of DO orders, there is not enough remaining to take care of civilian needs even at the reduced allotment.

As far as we can learn, there has not been a sufficient increase in the production of copper and zinc to satisfy this unprecedented demand and to take care of the requirement of the government stock piling, defense and foreign rearmament programs.

The high volume of production of brass mill products during 1950 and the early part of 1951, has been possible only because the mills have dipped into their metal inventories which are at their lowest ebb today.

From the brass mill standpoint, the short supply of zinc has been critical

because it is needed for alloying with copper to produce brass. Zinc acts as a means of stretching the available supply of copper because when alloyed with 100 pounds of copper it can result in about 145 pounds of brass. This increased tonnage will serve a larger number of users.

However, to augment the supply of zinc and copper, the brass mills use clean brass scrap, viz., clippings, rod ends, and shavings generated during metal goods fabrication. The scrap is added to the melt with new metal and is eventually turned into new mill products.

It has always been the custom for the brass mills to buy back the brass scrap from their mill products, customers and metal dealers at definite published prices which reflect the prevailing prices of new copper and new zinc.

Today, because of the voracious de-

mand for brass scrap, its price has been raised considerably beyond that of virgin metal. As a result, a portion of this scrap, which would ordinarily be returned to the brass mills, has been diverted to other channels.

The serious shortage of copper and zinc, plus the insufficient brass scraps being returned to the mills explains in part why the brass mills are operating on a reduced schedule and are not able to fill the many orders which have piled up.

Some Scrap Diverted

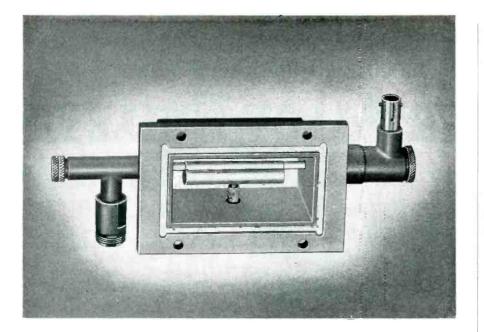
The scrap situation can be improved if metal goods manufacturers will make sure that *all* their scrap is returned to the brass mills. Although most manufacturers make it a practice to do so, there are some smaller ones who think that their scrap generation is insufficient to count. However, when the scrap generation of all small manufacturers are added up, the total becomes worthwhile.

It is true that in some cases brass scrap can fetch higher prices than that offered by the brass mills. Unfortunately, if this scrap does not find its way back to the brass mills, the brass mill tonnage is reduced and all brass users suffer.

How Brass Users Can Help

Here are a few suggestions which, if followed by metal goods manufacturers, would help the scrap situation considerably.

- Store all scrap in clean containers, keeping it separated by alloy and properly identified.
- When a sufficient quantity has been collected, notify the brass mill for whom it is intended. Give details as to weight, alloy, character of scrap. Ask the brass mill to collect it either in their own trucks or to designate the particular scrap dealer which does this work for the brass mill.
- If you know where there are clippings of zinc resulting from manufacturers' cups or stampings, or scrap copper wire or other forms of good clean scrap of brass, bronze or copper not generated by brass goods manufacturers, please notify your supplier of brass mill products. (6867)



Broad Band Mixer. . .

by Terpening

This mixer was designed by the Terpening engineering staff to meet a performance specification for low loss operation over a broad segment of the S Band. Local oscillator input and adjustable coupling probe are on the left; IF output at right. Important element is the cylindrical structure attached to the inner co-ax conductor within the waveguide section. This Terpening-developed element acts as a broad-banding transformer and permits high efficiency operation from 2700 to above 3000 mc. VSWR, with a JAN-approved crystal connected as shown, is less than 2.2/1 over the entire frequency range.

This is another example of the type of help Terpening is set up to provide prime contractors on microwave transmission line systems—from design through production. Although our engineering staff, laboratories and fully equipped shop are busy with government contracts, we will be happy to talk with you about your needs on similar work.



tions of any type. Consequently, satisfactory experience with this tube represents a considerable achievement as far as utilization of the new spacer material is concerned. Another interesting result was obtained when a number of terratex-spaced and mica-spaced 6J5-GT tubes were tested on a vibration-life test.

Figure 3 shows the remarkable difference in the vibration output characteristics of the two lots of tubes. The tubes were vibrated sinusoidally at 25 cps with an acceleration of $2\frac{1}{2}$ g and were read daily for a total vibration time of 400 hours. It can be seen that the terratex-spaced tubes had much lower vibration output than the mica-spaced tubes over the entire

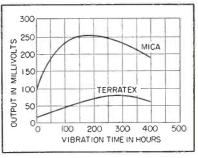
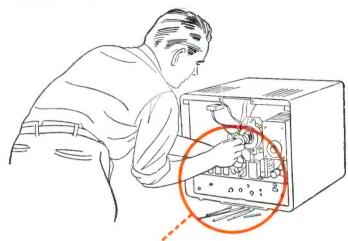


FIG. 3—Tubes employing terratex are less susceptible to microphonics

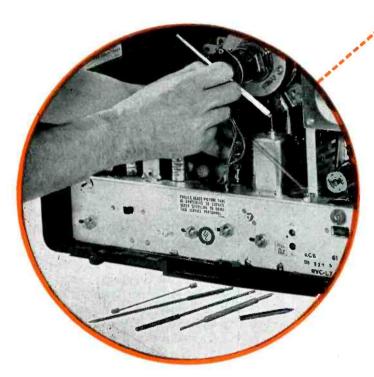
period of testing. Another particularly interesting characteristic of the material is the high surface roughness, which makes for good interelectrode insulation. Measurements of interelectrode leakage were made over a life-test period of 500 hours, and it was found that the terratex-spaced tubes were considerably superior to the micaspaced tubes at all testing intervals.

From this brief discussion of the material, it is apparent that there are several advantages associated with the use of terratex. From the standpoint of supply and fabrication, it is possible to obtain strips or sheets of terratex of uniform thickness made from materials available in this country. Further, the material has high thermal and surface resistivities and exhibits certain physical properties which make it appear to be superior to mica in some cases.

It is not known that terratex can be applied successfully as spacer material for all tube types, nor in



Flexible TV tools of Du Pont nylon plastic



TV aligning tools molded by Teal Molding Co., New Haven, Conn., for JFD Manufacturing Co., Inc., Brooklyn, N.Y.



are easy to handle... tough ... insulated against high voltage

Aligning the tuning circuit in television sets is a delicate job. It's precision work in hard-to-reach places within the set, working close to high-voltage equipment. These aligning tools made with nylon plastic make the job easier and safer.

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Demand for nylon plastic currently exceeds supply. However, we suggest you investigate its properties for future applications. We will gladly discuss the availability of experimental quantities for development work. For additional information on nylon and other Du Pont plastics, write:

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X1986 CERAMIC TERMINAL BOARD.

Grade L-5 silicone impregnated ceramic. 35/64" O.A. mounted height including terminals. 1½" length, ½" width. Assembly has 8 C.T.C. Type X1558 terminals in 2 rows (4 per row) 9/16" apart, plus two 4-40 tapped standoffs 3/16" high on ½" centers. A center ground strap is provided to which standoffs are riveted and soldered for good grounding at R.F. frequencies. All metal parts are non-ferrous, heavily plated.



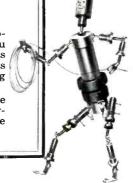


X1990 CERAMIC FEED-THROUGH INSULATOR. Grade L-5 silicone impregnated ceramic. $\frac{7}{8}$ " O.A. length including through terminal. $\frac{3}{8}$ " hex bushing threaded for $\frac{1}{4}$ " hole mounting. Voltage breakdown 4800 volts R.M.S. @ 60 cycles AC.

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its present development stage can it be used without decreasing exhaust machine production rates on certain tube types. However, upon the basis of the measurements which have been made on the material and upon the experience obtained in applying the processed terratex to vacuum tubes, it is likely that terratex could be applied to some types of tubes immediately that its several advantages over conventional mica spacers could be realized without sacrificing production rates or tube quality. In order to evaluate the material more fully and to improve it, work is being continued on the project at the G-E Owensboro Development Laboratory. Terratex processed for use in vacuum tube application is not commercially available at this time, although small samples for experimental use may be obtained through the Electronics Department of the General Electric Com-

This article is based on a paper presented by A. P. Haase and E. B. Fehr at the 1951 IRE Convention in New York.

Distributed Resistance Ignition Cable

FOR MANY YEARS it has been the practice to suppress radio frequency radiation from an automobile engine by the addition of resistance in series with the high-voltage ignition circuit. The addition of this resistance quickly damps out the radio-frequency oscillations set up by the discharge of the spark plugs.

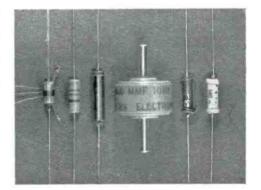
The most conventional form of resistor is a separate unit placed in series with the ignition leads either at the spark plug terminal or at the point where the high-voltage coil wire enters the center tower of the distributor. In some cases, resistances are built into the spark plugs. While each car presents a different problem of radio suppression, it is usually found that a 10,000-ohm resistor on the coil to distributor lead is all that is necessary to suppress radio interference to the point where ignition noise does not interfere with reception in the car radio.

With the advent of f-m and tele-



...if electronic components, such as molded coils are your problem

Other electronic components also built in quantity to your most exacting specifications for stability in service

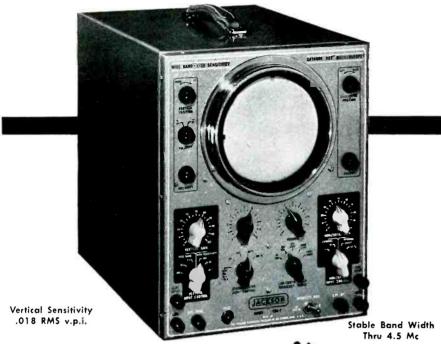




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Complete your Circuits with Resistors, Coil Forms and Iron Cores by Speer Resistor Corp., St. Marys, Pa. another SPEER CARBON CO. subsidiary



JACKSON Oscill gives you dual se

This is a high-quality, laboratory-grade 5" Oscilloscope that provides the "dual service" of both high sensitivity and wide band width.

s pecifications

Vertical Amplifier - Video-type frequency compensation provides flat response within 1.5 db from 20 cycles thru 4.5 Mc, dropping smoothly to a still useful value at 6 Mc.

Sensitivity Ranges — With a band width of 20 cycles thru 100 Kc, the sensitivity ranges are .018, 18, 1.8 RMS volts-per-inch. The wide band position 20 cycles thru 4.5 Mc has sensitivity ranges of .25, 2.5, 25 RMS volts-per-inch.

Horizontal Amplifier—Push-pull with sensitivity of 55 RMS volts-per-inch.

Input Impedances—Vertical. 1.5 megohms shunted by 20 mmfd. Direct to plates, balanced 6 megohms shunted by 11 mmfd. Horizontal: 1.1 megohms

Linear Sweep Oscillator - Saw tooth wave, 20 cycles to 50 Kc in 5 steps. 60 cycle sine wave also available, as well as provision for using external sweep.

Input Voltage Calibration—Provides a standard voltage against which to measure voltages of signal applied to vertical input.

Vertical Polarity Reversal—For reversing polarity of voltage being checked or for choosing either positive or negative sync. voltages.

Return Trace Blanking—Electronic blanking provides clear, sharp trace to prevent confusion in waveform analysis.

Synchronizing Input Control-To choose among INTERNAL, EXTERNAL, 60 CY-

CLE, or 120 CYCLE positions.

Intensity Modulation—60 cycle internal or provision for external voltage for intensity modulation uses.

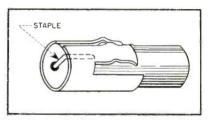
Additional Features—Removable calibration screen—Accessory Model CR-P Demodulation Probe for Signal Tracing—Allsteel, gray Ham-R-Tex cabinet. Total net weight only 26 pounds. Same height as other Jackson TV instruments: 13" H x 101/4" W x 151/8" D.

Prices: Model CRO-2, Users' Net \$197.50.

Model CR-P Probe, Users' Net \$9.95.

vision, the interference from automobile engines became more seri-Most cars, even those suppressed for ordinary radio, constitute a public nuisance and interfere with reception when operated in close proximity to tv sets.

It has been found by engineers at the Packard Electric Division of General Motor Corporation that the use of suppressor ignition cable having a distributed resistance of approximately 5,000 ohms per foot completely eliminates this objectionable interference. The conductor of this cable consists of multiple threads of high-strength materials, such as linen and rayon. enclosed in a rayon braid. The desired conductivity is imparted to this structure by impregnation with graphite solutions and treatment with special conducting coatings so



Distributed-resistance cable used in connecting automobile spark plugs reduces f-m and television interference

that cable insulation will properly adhere to the conductor.

Special techniques are employed in attaching terminals to this high resistance conductor. It has been found that a piece of metal wire approximately & in. long must be inserted in the end of the conductor in order to provide a well-distributed area of contact with the fabric conductor, as shown in the accompanying diagram. An extension of this metal pin is carried across the end of the cable and is firmly contacted by a conventional terminal.

Extensive field and laboratory endurance tests indicate that this method of connection will stand up satisfactorily under even the most rigorous service conditions.

Tests showed the high-resistance cable and the standard neoprene ignition cable tested equivalent on all tests except the voltage breakdown on the new cable is approximately 5 kv less than standard

TWO OTHER FINE JACKSON INSTRUMENTS

Model 655 Audio Oscillator

Model TVG-2 TV Generator



Addio Ostinu.

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200,000 cycles, Less than
5% harmonic distortion between 30 cycles and 15,000
cycles. Frequency calibration accurate within 3% or
1 cycle, Hum level down
more than 60 db of maximure dances of 10 250,
200,000 cycles of the 10 250,
200,000 cycles
200,0

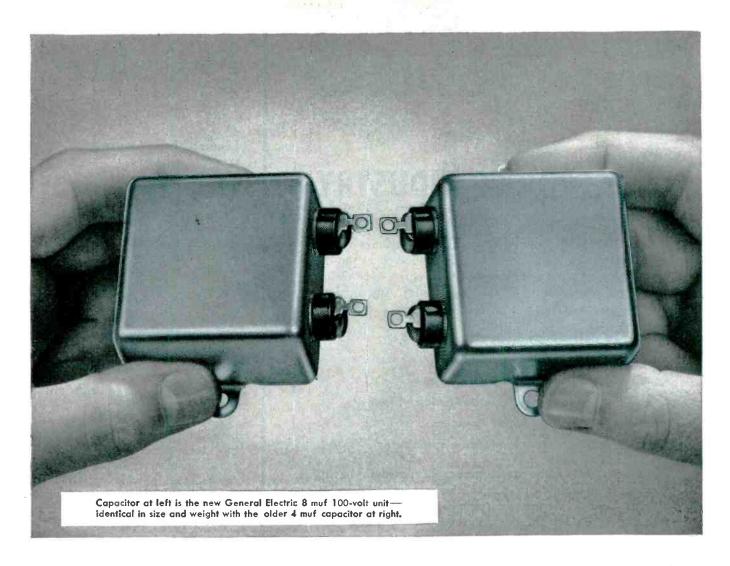
Model IVG-2 IV
Sweep Oscillator in three
ranges from 2 Mc thru 216
Mc. all on fundamentals.
Reversible sweep direction.
Sweep width variable -1 Mc
thru 6 Mc. Marker revenue
thru 6 Mc. Marker revenue
thru 6 Mc. Marker Covenie
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8 muf...in the space of 4

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Here is another outstanding G-E capacitor development—thinner paper, thinner foil, so that double or triple the capacitance can be designed into a cubic inch.

These new capacitors are comparable in all ways with previously offered paper dielectric units, are equally dependable, and in addition are smaller in size and lighter in weight. They will not introduce noises into the system. They will satisfactorily pass signal voltages approaching zero. Their insulation resistance values remain high after long periods of service. While primarily intended for d-c applications with allowable ripple voltages in accordance with JAN-C-25, they will withstand occasional discharges, and can be used in low-voltage a-c circuits.

In Regular Production. Units of 3, 8 and 10 muf in Case Style 53 and 4 muf in Case Style 61 are in regular production. Other ratings can be built in mass-production quantities.

These capacitors meet all requirements of "F" characteristics of JAN-C-25 for 100-volt d-c units. For applications where an expected life of 1000 hours at 40 C is satisfactory, rating can be increased to 150 volts. For ambients above 40 C, units should be derated in accordance with JAN-C-25 Specifications. There is negligible change in capacitance from -40 C to 105 C—and units will give full life at temperatures as low as +55 C.

If you have applications involving reasonable quantities, get in touch with us. Your letter, addressed to Capacitor Sales Division, 42—304, General Electric Company, Pittsfield, Mass. will receive prompt attention.

General Electric Company, Schenectady 5, N. Y.



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In addition to the development of powders having specified magnetic, or electrical characteristics, PLASTIC METALS offers iron and nickel powders which are being used by the electronics field in the form of parts made by powder metallurgical methods. The present shortages of steel, nickel, aluminum, zinc, etc. which aggravate the problems of production, or procurement of such things as castings, forgings and die-castings, make it important that you investigate the possibilities of powder metallurgy for:

- 1. Conservation of critical metals
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- 3. Improvement in product quality
- 4. Attainment of greater output

PLASTIC METALS invites you to submit your problems that might involve iron, nickel, manganese, silicon or magnetic iron oxide powders. Our Research Staff and 17 years of experience are at your disposal.

For more detailed information, write to



cable, also the capacitance is 5 $\mu\mu f$ per foot greater. The high resistance cable is satisfactory for automotive ignition.

Logarithmic Counting Rate Meter

By W. F. GOODYEAR

Tracerlab. Inc.

Boston. Massachusetts

Counting rate meters in conjunction with chart records have found wide application in laboratories where measurements of radioactivity are being conducted. They possess distinct advantages over scaling type instruments in that counting rate can be observed directly without computations and that with a time record inconsistencies in statistical variations can be instantly spotted and either accounted for or disregarded.

If the rate-meter response can be made logarithmic, its utility can be further extended. As an unattended monitoring device it is ideal for measuring wide variations in rate with precisions that are unattainable in a linear instrument without range switching.

Figure 1 shows the front panel of such an instrument designed with the logarithmic response in the circuit itself, permitting the use of conventional panel meters and external linear recorders. The logarithmic element is a thermal diode operating under retarding plate conditions. The operation of such tubes has been described elsewhere, 1,2,8 and where the leakage resistance can be made sufficiently large, they maintain their logarithmic response over extremely wide ranges of plate current.

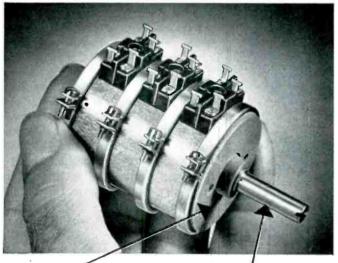
In this circuit, intended to operate from the a-c line, heater-to-

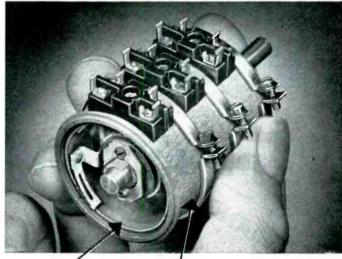


FIG. 1—Front view of logarithmic counting rate meter

How do you put PRECISION into a potentiometer?...

... At Fairchild we do it by careful design plus scrupulous attention to every phase of manufacture from the selection of raw materials to the precision machining of all mechanical components. Every mechanical part that affects the potentiometer's accuracy is made in our own plant. The result? Let's find out by examining one of our new 746 units for precision.





PRECISION in the shaft. The shaft is centerless-ground from stainless steel to a tolerance of +.0000, -.0002 in. which, together with the precision-bored bearings, results in radial shaft play of less than .0009 in. These accuracies are essential to the utilization of the full precision of the windings and associated gearing.

PRECISION in the housing. Precision-machined out of solid aluminum bar stock, every housing is perfectly cylindrical—and stays that way—resulting in better accuracy of potentiometer output. This construction also results in perfect fit and alignment between housings to permit ganging of up to 20 units on a single shaft without any eccentricity of the center cups even though only two bearings are used for the entire gang.

PRECISION in the mounting plate. All critical surfaces of the mounting plate are accurately machined at one setting to insure a shaft-to-mounting surface squareness within .001 in./inch and concentricity of shaft to pilot bushing within .001 in. FIR. Such precision is essential in order to eliminate backlash or binding in the precision gearing used to drive these potentiometers.

PRECISION in the windings. The windings of all Fairchild precision potentiometers are custom-made by an exclusive technique on machines designed and built by Fairchild expressly for the purpose. Guaranteed accuracy of linear windings in this potentiometer is 0.5%; non-linear, 1.0%. Higher accuracies available in other types. Guaranteed service life—1,000,000 cycles.

Do your requirements call for this kind of precision? If you have a special problem, Fairchild Potentiometer Sample Laboratory engineers are always available to help you. For complete data, write, stating your requirements, to Fairchild Camera and Instrument Corporation, 88-06 Van Wyck Boulevard, Jamaica 1, N. Y. Dept. 140-17A.





cathode leakage was largely the limiting factor at the lowest current ranges. The results of a small portion of a group of unselected 6AL5 diodes operated in this fashion are shown in Fig. 2. The curves are displaced horizontally for clarity in showing consistency of slope. If not displaced, all of the curves would very closely coincide with each other giving a resultant slope of about 200 millivolts per decade of current.

At 10⁻⁰ amperes, the leakage resistance is just becoming appreci-Actually the tubes were measured down to 10-10 amperes at which point the leakage is up to 100 percent of the signal current. As shown in Fig. 4, the logarithmic diode V_{1A} is operated with its cathode at +105 volts, which, with a 10⁻¹⁰ ampere limit on leakage, requires a resistance of at least 10-12 ohms to ground. Since heater-tocathode leakage is far in excess of this figure, a special heater transformer was constructed in which the secondary was wound separately and wrapped with Teflon tape before assembly to ensure a high d-c resistance.

At 10⁻⁴ amperes, five of the six diodes shown in Fig. 2 have a departure from the ideal slope of less than 25 percent. Assuming that the departure is a linear error at both the high and low current limits, it may be predicted that these errors will be less than 2.5 percent for the current range of

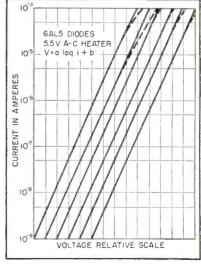
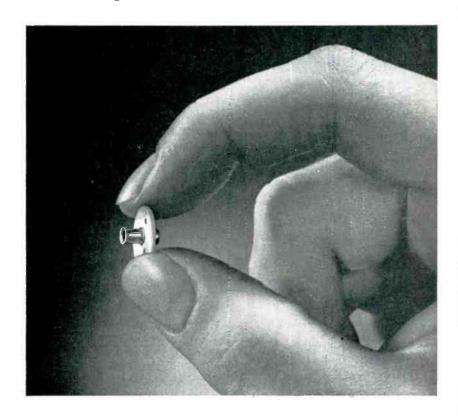


FIG. 2—Results of small sample of random 6AL5 diodes. Curves are displaced horizontally for clarity

Disc Cathode Speeds Assembly-Improves Performance



• Electronics manufacturers find it pays to be a customer of Superior. They receive good service, quality products and the benefits of Superior's methods and metals research that constantly improves upon already good products.

An example is the new, improved Disc Cathode. Investigation proved that a slight flaring of the open end minimized the danger of heater cathode "shorts" caused by scraping of the heater wire coating during insertion, while speeding the operation.

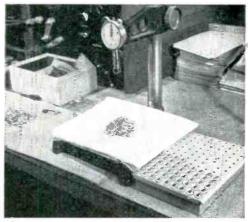
This feature added to an already excellent cathode, resulted in a

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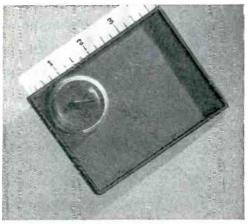
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10⁻⁵ to 10⁻⁸ amperes. Subsequent tests have verified this and later tests on the Raytheon CK5829 subminiature equivalent of the 6AL5 have indicated this type is even more consistent and more linear at the higher currents, permitting errors of less than 1 percent for the range just mentioned.

With this accuracy over three decades, a Brown type potentiometer recorder with its wide chart was selected to enable an operator to utilize fully the inherent instrument accuracy. For the same reason a panel meter with a 270-degree movement was selected for operation in the absence of a recorder.

The counting rate range was chosen to be from 1 count per second to 1,000 counts per second for the three most useful decades of operation. Conversion of counting rate to average diode current was accomplished in a circuit shown simplified in Fig. 3.

The switch is closed for a fixed duration t for each individual impulse. A large capacitor shunts the diode to keep its average voltage from changing to any extent during individual impulses so that on a peak, current E/R flows for each input signal if E is much larger than the diode voltage. From this the average current can be readily calculated.

The diode response has been given as

$$V = \frac{KT}{a} \log i + b \tag{1}$$

where K is Boltzmann's constant, e is the electronic charge, and T is the cathode temperature in de-

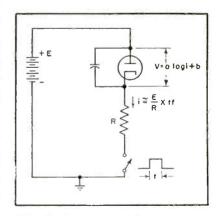
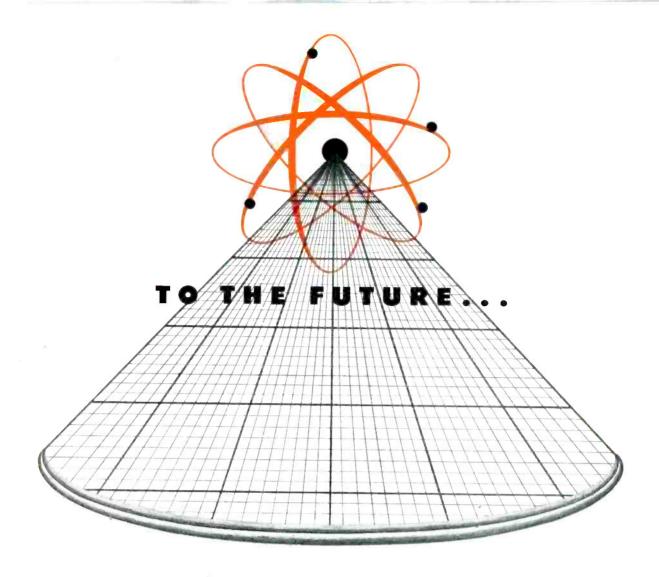


FIG. 3—Basic circuit for converting counting rate to average diode current



ELECTRONIC RESEARCH AND DEVELOPMENT

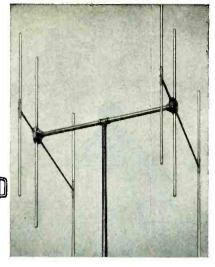
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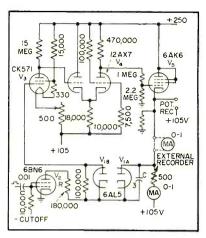


FIG. 4—Complete schematic of logarithmic counting rate meter

grees Kelvin. Differentiating,

$$\frac{dV}{di} = \frac{KT/e}{i} \tag{2}$$

which for conventionally operated oxide coated cathodes becomes simply

$$\frac{dV}{di} = \frac{0.1}{i}$$

This is recognizable as the dynamic resistance of the diode and indicates that the resistance will vary inversely with counting rate.

This resistance in parallel with the tank capacitor forms the circuit time constant

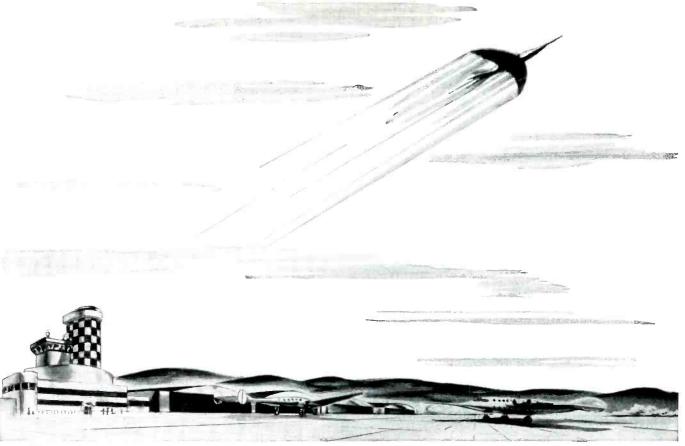
$$RC = \frac{0.1C}{i}$$

Unlike a conventional ratemeter, the time constant is not fixed but varies with counting rate. Since the fractional error in counting rate at any given instant is

$$E = 0.67 - \frac{1}{\sqrt{2n RC}} \tag{4}$$

where n is the number of pulses in the time RC, the fractional error in this type of circuit tends to remain constant for a wide range of actual counting rates. Also, because of the progressive resistance change this logarithmic circuit will respond more quickly to large changes in counting rate and reach equilibrium faster than a linear circuit with the same time constant for the initial rates.

Figure 4 shows the voltmeter and charging circuits designed for this circuit. The 6BN6 (V_2) is used as the switching device in series with the current limiting resistor and the diode. Stringent demands



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are placed on this switching tube. During conduction, the tube must pass peak currents of about 10-8 amperes, yet between conducting intervals its plate current due to all causes including leakage must be less than 10⁻¹⁰ amperes to permit 1 percent accuracy at 10-8 amp diode current. Many tube types were tested for this application with the final selection being the recently announced beam type 6BN6 designed as an f-m discriminator-limiter. The full conduction current is relatively independent of control pulse amplitude and its leakage current is considerably better than the 10⁻¹⁰ amperes required. In fact this tube's performance under these somewhat unique conditions is far superior and far more consistent than with any other conventional type tested.

Voltmeter Circuit

The voltmeter circuit is of the feedback amplifier type consisting of V_3 , V_+ and V_5 . The grid of V_3 sees the sum of the diode voltage, applied through V_{1B} in series with the feedback voltage developed across the sensitivity control and milliammeter. As a result the voltage from the cathode of the logarthmic diode to ground is maintained nearly constant and since the feedback voltage must nearly equal the signal voltage, the current required to supply the feedback voltage depends on the amount of resistance in the sensitivity control. An external recording milliammeter can be also connected in the feedback loop since current is supplied, or a potentiometer type recorder can read the feedback voltage.

Tube V_3 is an electrometer-type tube with its d-c heater operated from the regulated B supply voltage. The zero control is in the heater circuit and operates by changing the relative grid bias. Tube V_{1B} is the unused section of the logarithmic diode and is connected to help buck out the effects of line voltage changes on V_{14} .

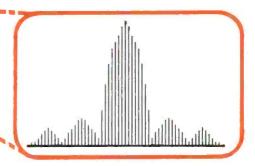
Despite this latter precaution, for low long-term drifts, it is also necessary to regulate the line voltage supply to the instrument. Likewise, B supply regulation is essen-

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ment is difficult without two accurrately known rates spaced far enough apart to adequately define the slope. In this case a decade scale of ten was employed to divide the line frequency to the input pulse forming circuits preceding the charging tube. Sixty cycles was used directly for the other calibration point. The plug-in decade scale can be seen in Fig. 5. It is also used to divide the signal rate if desirable, shifting the 1 to 1,-000 count per second range by a factor of 10. Figure 5 also shows the large tank capacitor mounted



FIG. 5-Rear view of unit shows plug-in decade unit and large tank capacitor (directly behind speaker)

directly beneath the speaker. It is the polystyrene dielectric type and is used to minimize hysteresis effects due to dielectric absorption and because of its high leakage re-

The remaining circuitry is fairly conventional. A regulated highvoltage supply is provided for Geiger counters or similar-type radiation detectors. The output of the detector is amplified and shaped to give constant width pulses in the input circuits. An aural monitor has also been included to give a rough indication of counting rate at very low levels.

REFERENCES

(1) R. E. Meagher and E. P. Bentley, Rev. Sci. Inst., Nov. 1939.
(2) T. S. Gray and H. B. Frey, Rev. Sci. Inst., Feb. 1951.
(3) "Waveforms," Vol. 19, Rad. Lab. Series, p 61, McGraw-Hill Book Co., Inc., New York.
(4) W. C. Elmore and M. Sands, "Electronic Techniques," p 252, National Nuclear Energy Series.



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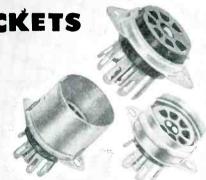
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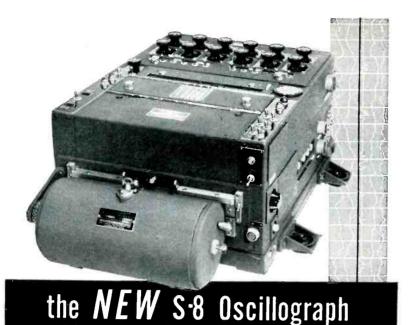
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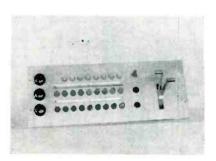
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(continued from page 144)

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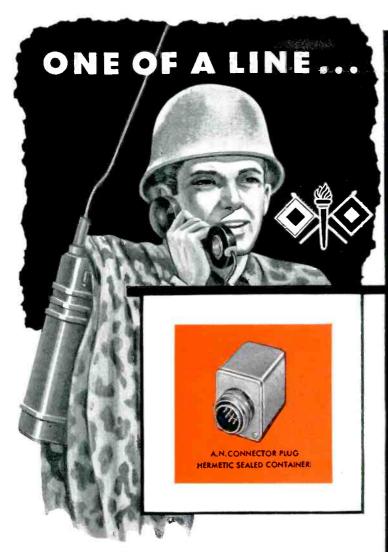
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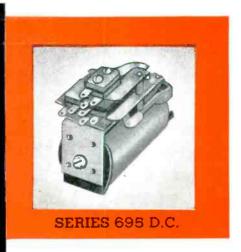
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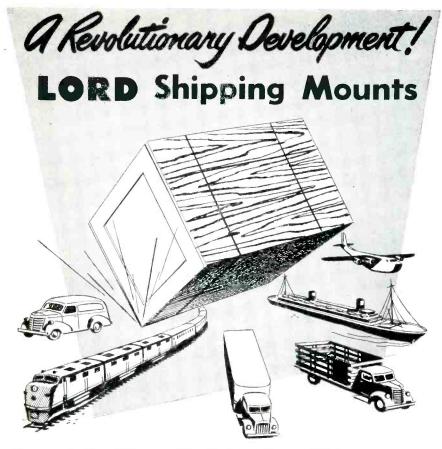


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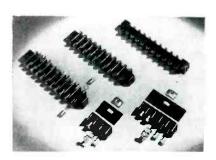
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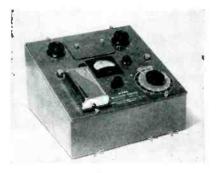


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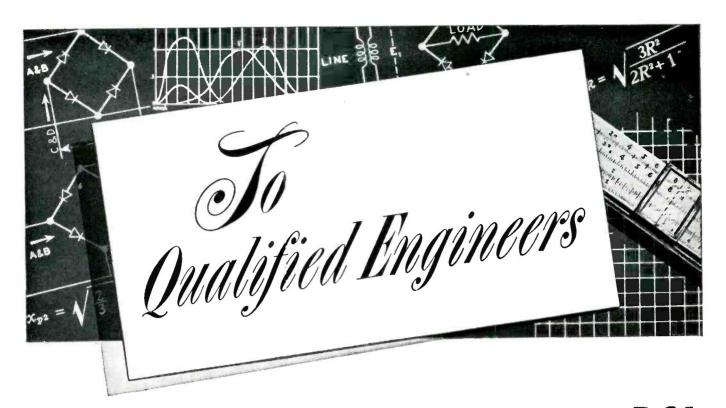
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If you qualify for any of the posit listed here, by all means write us a personal interview—include y résumé. Write to: Mr. Robert McQuiston, Specialized Employment Division, Dept. S-87, Radio Corporation of America, 30 Rockefeller Plaza, New York 20, N.Y.



PADIO CORPORATION of AMERICA



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UP TO 4-POLE, DOUBLE THROW

Wherever severe shock and vibration are encountered, this rugged little relay can be relied upon to give dependable service. Originally designed for aircraft and mobile equipment, the versatile AMRECON Type DO is adaptable to many other uses where maximum protection against mechanical injury is required. Small in size (2½" high, 2½" wide, 1½" deep) it is available in contact arrangements up to four-pole, double throw. Coils are normally rated at 3 watts d.c., or 6 watts, 60-cycle a.c.—for voltages up to 230 volts d.c., or 440 volts a.c. Contact rating: 10 annps—at 115 volts a.c. non-inductive, or 32 volts d.c. Weight: approximately six ounces.





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ville, Tenn. Model 144 megohm bridge is a Wheatstone bridge specially designed for measurement of high resistances. Resistances of 10° to 10° ohms can be measured. Special insulation, shielding and an electrometer type detector are incorporated to permit measurements in this range. A set of resistors calibrated by the National Bureau of Standards is supplied with each instrument. The bridge operates from self-contained batteries.



Linear Potentiometer

CHICAGO INDUSTRIAL INSTRUMENT Co., 536 W. Elm St., Chicago 10, Ill., has introduced the Selectohm, a new accurately calibrated 0 to 100,000-ohm linear potentiometer for use as a resistance substitute for service or laboratory work. It serves as a precision rheostat, shunt or multiplier, and is rated at 25 watts. Price is \$7.50.

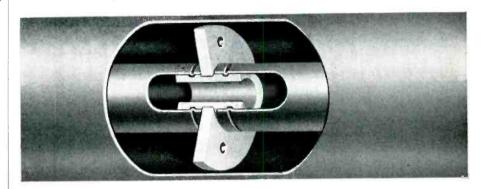


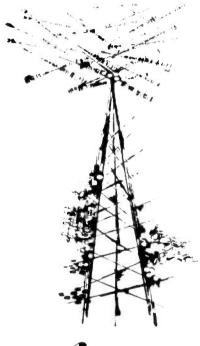
Frequency Standard

AMERICAN TIME PRODUCTS, INC., 580 Fifth Ave., New York 19, N. Y. Type 2005 frequency standard is a

Andrew VHF and UHF "TEFLON" insulated

TV TRANSMISSION LINES





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Teflon insulators, with a dielectric constant of 2.0, ½ that of steatite and a loss factor of 0.0004, ½ that of steatite—minimize impedance discontinuity, increase efficiency . . . Andrew further compensates for insulators in 3½" and 6½" line as illustrated—on smaller diameters, insulators are secured in a rolled groove on the inner conductor. A complete line of hangers, elbows, gas barriers and other accessories are available.

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Andrew Bulletin 73—to help you select correct transmission line for your TV station—a complete table of power ratings and loss data for use over entire UHF and VHF TV bands— No obligation—write for your copy today!

ANDREW CORPORATION 363 E. 75th Street, Chicago 19 Please send me a copy of Bulletin 73 describing VHF and UHF Television Transmission lines.				
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The required resistor is an integral part of this assembly —"built-in."

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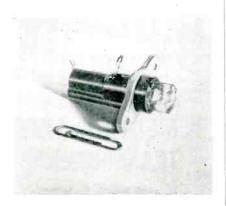
The only seals you can hot tin dip at 525°F. for easy assembly soldering, for a strain and fissure-free sealed part with resistance of over 10,000 megohms!

Hermetic headers withstand high vacuum, high pressure, temperature cycling, salt water immersion and spray etc., and are used extensively by America's leading industries and government agencies.



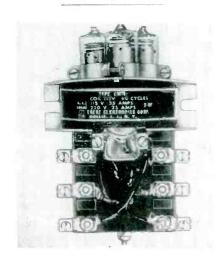
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useful laboratory tool, suitable for running gyros, small motors, controlling inverters, stroboscopes and the like. Input power is 45 watts at 110 volts from 50 to 500 cycles. Output power is 10 watts at 115 volts at any specified frequency from 50 to 500 cycles. Accuracy is ± 0.001 percent in room temperature range. It measures 8 in. \times 8 in. \times $7\frac{1}{4}$ in. high.



Indicating Fuseholder

ALDEN PRODUCTS Co., 117 N. Main St., Brockton 64, Mass., has developed model 440–3FH miniaturized indicating fuseholder that instantly spots a blown fuse. Its design features a neon bulb and double contacts molded as an integral part of the crystal-clear lens. The unit is ideally suited to mass production assembly techniques. It has easily accessible solder tabs that facilitate wiring and assembly to other circuit elements.



Three-Pole Relay

EBERT ELECTRONICS CORP., 185-09 Jamaica Ave., Hollis 7, N. Y. Type



The largest commercially available equipment for shake testing according to military specifications is now in service at Sperry Gyroscope Company. Developed by Calidyne to deliver a force output of 2500 pounds at frequencies up to 500 cycles per second, this electrodynamic shaker, with its associated power supply and control system, is the latest addition to Calidyne's complete line of equipment for vibration studies.

Further original Calidyne developments in the field of vibration investigation include other electro-dynamic shakers, vibration pick-ups, couplers, vibration standards, vibration meters, and calibrators for accelerometers and vibration pick-ups. Each of these was produced to satisfy a recognized need in vibration research and the advanced thinking they demonstrate has earned their extensive use in diverse fields of engineering.

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Hermetically sealed type transformers to specifications.

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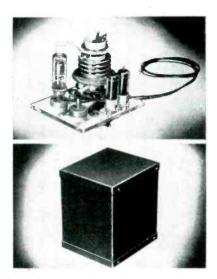
Transformers up to 225 KVA are built with attention to detail perfection to assure high performance, long-life service.

ACME ELECTRIC CORP.

Cuba, N. Y., U. S. A.



Em-7 three-pole relay employs one coil and its tungsten contacts are hermetically sealed in hydrogen-filled glass tubes. This construction provides a reliable and positive simultaneous contacting means for loads up to 35 amperes at 115 volts a-c, or 25 amperes at 220 volts a-c per contact. These relays can be supplied with all three contacts either normally open or normally closed, or with any combination of normally open and normally closed contacts. Dimensions are $4\frac{1}{16}$ in. wide, $5\frac{5}{8}$ in. tall by 3 in. deep.

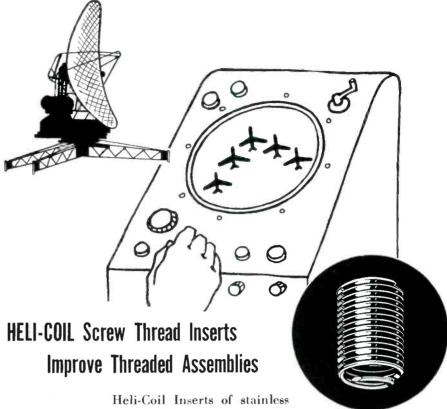


D-C Power Supply

Spellman Television Corp., 3029 Webster Ave., New York 67, N. Y., has announced a new compact high-voltage unit designed for industrial or laboratory applications. It is a well-filtered 4,500 volt r-f d-c power supply, completely enclosed measuring only $5\frac{s}{2}$ in. \times $4\frac{1}{2}$ \times 5 in. Voltages up to 7,500 volts at 1 ma may be obtained. It is available with either negative or positive polarity outputs.

ULF Oscillator

KROHN-HITE INSTRUMENT Co., 580 Massachusetts Ave., Cambridge 39, Mass. Model 410-A ultralow frequency oscillator covers the subaudio and the entire audio range. It provides both sine wave and square wave at any frequency range between 0.02 and 20,000 cps. The unit is ideal for medical research,



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can provide an armored thread in every tapped hole, in any material . . . eliminate wear, add strength, prevent stripping and seizing, resist corrosion and galling. Valuable savings in weight and space are possible where these precision-formed wire inserts are used.

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Literature____

Data Bulletins. International Resistance Co., 401 N. Broad St., Philadelphia 8, Pa., has available 7 catalog data bulletins dealing with the following: type MP high-frequency resistors; type MV high-voltage resistors; type BW insulated wire-wound resistors; type UHR ultrahigh wire-wound resistors; type UHR ultrahigh resistance resistor; a line of precision wire-wound resistors; the 1101 general-purpose digital computer; and the type W 2-watt rheostat potentiometer. Complete technical information and illustrations are included.

Autographic Recorder. Baldwin-Lima-Hamilton Corp., Philadelphia 42, Pa., announces the 4-page bulletin 330 describing and illustrating the model MD-2 autograph recorder. The instrument described is a portable, self-contained Microformer type that will plot automatically on rectangular coordinates any two variables that can actuate the movable cores of miniature variable transformers, either directly or through accessory Bourdon tube units.

Cellular Rubber. The Sponge Rubber Products Co., Derby Place, Shelton, Conn. A 20-page booklet deals with the properties of test data on and specifications for cellular rubber. The product is useful as a turntable cushion to prevent transmission of motor vibrations to recording discs; as an amplifier insulator to provide a wall of insulation between plug-in amplifier and main electrical chassis; and as hand microphone spacers to protect the element from damage and insulate it against transmitting the sound that would oth-

There's an GPotentiometer for every application

Miniaturization of precison potentiometers is keeping pace with the increased demand for smaller assemblies and compact design. Now you can minimize wasted space with TIC's outstanding, new RV ½* and RV 1-1/6* Miniature Potentiometers.

In spite of their thumbnail size the RV ½, and the RV 1-1/16 are precision, high linearity variable resistors, (or adjustable trimmers) of high stability—achieving a standard of performance hitherto unavailable in such miniature potentiometers.

Construction features include: precision machined aluminum base ... low torque ... all soldered connections

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specifications. Ganging if desired with the adjustation clamp ring.

NY 7/3* available with linear resistance elements only—nine standard resistance values from 100 to 25,000 ohms.

Power rating 6 watts at 25°C. Illustration shows RV 7/3 with threaded bushing . . . servo mounting available if desired.

RV 1-1/6* available with linear or non-linear resistance elements—nine standard resistance values from 100 to 50,000 ohms. Illustration shows RV 1-1/16 with 3 tapped hole mounting . . . servo mounting or threaded bushing if desired.





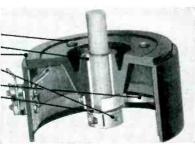
Type RV1%* and RV2* High Precision . semi-standardized types Potentiometers . . of precision machined aluminum base potentiometers with exceptionally high electrical accuracy and mechanical precision. For both linear and non-linear functions. Designed for precision instrument, computer and military applications. Accurate phasing of individual units possible with clamp-ring method of ganging.

*Numbers refer to diameter of bases.

Ball bearing models available.

Tapped mounting inserts Bronze bushing Totally enclosed with cover "Constrict-O-Grip" clamping to shaft -(no set screws) Precious metal contacts Silver overlay on rotor take-off slip ring

Sine and cosine potentiometers available in RVP3* and RV2* bases.



Type RV3* Bakelite Base Precision Potentiometers . . . available in models for either linear or non-linear functions. Stock resistance values ranging from 100 \$\Omega\$ to 200,000 \$\Omega\$ and power ratings of 8 and 12 watts. 360° mechanical rotation or limited by stops as desired. Potentiometers of this type available to widely varying accuracy requirements (linearity to ±0.25%) see TIC Bulletin RV3-250. Special models available for high humidity applications.

Type RVP3* High Precision machined aluminum base Potentiometers . . . available in models for either linear or non-linear functions with standard resistance values up to 200,000Ω. Linearity to $\pm 0.1\%$. Eleven gang assembly available, example of TIC's potentiometers multi-ganged with TIC's adjustable clamp ring. Can be supplied to meet various mounting requirements - single hole, 3 tapped hole mounting or servo mounting as desired.

RVT Translatory **Potentiometers** . actuated by longitudinal instead of rotary motion providing linear electrical output proportional to shaft displacement. Used as a position indicator, high amplitude displacement type pickup and for studying low frequency motion or vibration. Features exceptionally high linearity and resolution. Available in various lengths and resistance values.





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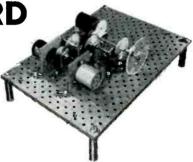
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If this sounds like your kind of opportunity, we'd be glad to tell you more about it-and about the fine living conditions and salaries in Minneapolis and Philadelphia. Depending on the location you prefer, write to H. D. Elverum, Personnel Department EL-1, Minneapolis 8, Minn., or W. Reiterman, Personnel Department EL-1, Philadelphia 44, Pa., giving your qualifications and experience. Your correspondence will be held in the strictest confidence, of course.

Honeywell

First in Controls

erwise result from dropping or banging the microphone.

Connectors. American Phenolic Corp., 1830 South 54th Ave., Chicago, Ill. Catalog A2 contains information for selecting and ordering AN connectors for power, signal and control circuits in aircraft and electronic equipment. The different types of connector shells, insert arrangements and fittings are illustrated. Complete listings of the thousands of arrangements are given. Dimensions, weights and electrical data are shown in detail.

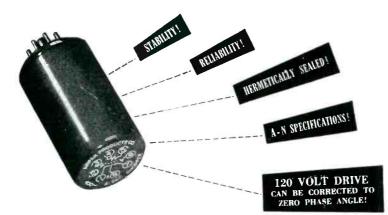
Vibration Isolators. The Barry Corp., 700 Pleasant St., Watertown, Mass. Catalog 509 discusses the structural features of all-metal vibration isolators that employ no organic material and are unaffected by extreme temperatures. Unit mountings, for direct attachment to equipment, and standard size mounting bases are described. The mountings covered meet the requirements of JAN-C-172A, MIL-E 5272 and MIL-T-5422. The publication lists load ranges and dimensions of both unit and mounting base types.

Direct-Writing Recorder Amplifier. Baldwin-Lima-Hamilton Corp., Philadelphia 42, Pa., has published bulletin 332 covering the model 141 direct-writing recorderamplifier assembly. The instrument described is designed for inkless recording of electrical unbalance of a resistance bridge when using strain sensitive elements such as SR-4 strain gages or temperature-sensitive elements. The bulletin deals with performance and operation of amplifier and recorder assemblies, and lists available accessories.

Miniature Standard Cell. Muirhead & Co., Ltd., Beckenham, Kent, England, has issued a 2-page bulletin dealing with the type D-550-B miniature standard cell that has been produced especially for use in educational establishments where it is desirable that its internal construction should be visible so that its working may be more easily understood. Specifi-



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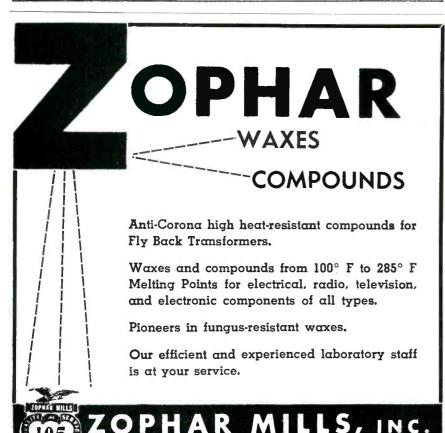


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400 CYCLE CHOPPERS

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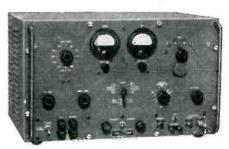


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Standard signal source for complete testing of VHF airborne omnirange and localizer receivers in aircraft or on the bench is ARC's Type H-14 Signal Generator. It checks up to 24 omni courses, omni course sensitivity, to-from and flag-alarm operation, left-center-right on 90/150 cycle and phase-localizers, and all necessary quantitative bench tests. Permits quick, accurate, check-out of aircraft just before take-off. For ramp checks RF output 1 volt into 52 ohm line; for bench checks, 0-10,000 microvolts. AF output available for bench maintenance and trouble shooting.

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Type H-12 VHF Signal Generator 900 - 2100 mc - source of cw or pulse amplitude-modulated RF. Power level 0 to -120 dbm. Internal pulse circuits with controls for width, delay, and rate, and provision for external pulsing. Frequency calibration better than 1%. Built to Navy specs for research, production testing. Equal to

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AIRCRAFT RADIO CORPORATION

Boonton **New Jersey** Dependable Electronic Equipment Since 1928 cations given include general description, electrical characteristics, current discharge rate and dimensions.

Pneumatic Tools. Newage International, Inc., 235 E. 42nd St., New York, N. Y. Catalog PT deals with a complete line of pneumatic tools including drills, screwdrivers and nut runners, countersinking tools, shears, nibblers, grinders and accessories. An illustrated description of each tool is given.

Tube Catalog. General Electric Co., Syracuse, N. Y. A new 107-page pocket-size handbook lists the essential characteristics of every tube likely to be found in any a-m, f-m or tv home receiver. The reference contains ratings and other data essential to fast, efficient trouble shooting. Basing diagrams for each of the 856 different tube types listed are shown on the page with the data. Price is 35 cents.

Electrical Insulation. The INSL-X Co., Inc., Water St., Ossining, N. Y., has issued a booklet on its electrical insulating and specialty compounds. Among the many uses of the basic raw material covered are fungicidal coatings, arcresisting coatings, coil coatings, tool insulation and insulating varnishes. Complete technical description, curing-time-hours chart and price list are included.

Microwave Test Equipment. Kings Microwave Co., Inc., 50 Marbledale Rd., Tuckahoe 7, N. Y. A recent 24-page booklet deals with a line of microwave test equipment and components. Included are illustrated descriptions of highpower terminations, tees, water loads, low-power waveguide terminations, glide screw tuners, flap attenuators, bidirectional couplers, micrometer wavemeters, waveguide chokes and flanges and similar equipment.

Silvered Mica Capacitor. Cornell-Dubilier Electric Corp., South Plainfield, N. J. Bulletin NB 141 announces the type 22R high-stability silvered mica capacitor in

High Temperatures? Small Space? Severe Conditions?



Today's trend towards miniaturization makes severe demands on electronic wiring. That is why so many leading electronic equipment manufacturers specify Rome Synthinol 901. A resin plasticized polyvinyl chloride thermoplastic compound, it is Underwriters' approved for temperatures up to 105° C., as a special small-diameter type with no assigned voltage, as well as for regular 300 and 600 volt ratings.

Where space is a problem, Rome Cable engineers have developed a small-size hook-up with with an 8 mil wall of Rome Synthinol 901 insulation and an exceptionally small over-all diameter. With nylon sheath maximum diameters run from as small as .051" for 24 AWG to .100" for 16 AWG.

What's more Rome Synthinol 901 has greater resistance to heat deformation, baking embrittlement, shrinkage, cracking, solvents, moisture and flame, plus improved solderability. It is available in distinct, permanent colors, either plain or with outer coverings.

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Rome Cable is an approved manufacturer of military types SRIR, SRHV and WL, complying with Army-Navy Joint Specification JAN-C-76, as well as shipboard types SRI and SRIB conforming to Specification MIL-C-915. Insulated with Rome Synthinol thermoplastic compound, these wires are manufactured in the complete range of specification sizes.

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Any type of condenser... paper, mica, oil filled, ceramic or electrolytic... can be graded on the PC-4 at rates up to 8000 per day by an unskilled operator. Working to an accuracy of 0.2%, the PC-4 is a companion production instrument to the famous PR-5 Automatic Resistance Comparator. Leading manufacturers have found it an indispensible tool in the fight for higher quality and lower production costs. Easy operation reduces inspection time to an absolute minimum.

Completely self-contained, the PC-4 requires no outside attach-

ments other than the Standard Capacitor against which the unknowns are to be checked. Operates on 110 Volt—60 cycle AC. Range: 10 mmfd to 1000 mfd. Size: 18" x 12" x 12". Weight: approximately 35 lbs. For complete details, write for Catalog Sheet 7-E.

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NEWS OF THE INDUSTRY

(continued from page 148)

ton cargo vessel is at least \$2,000 a day.

Equipment to be used until the close of the experiment in December, 1951, has been furnished without cost by Sperry and Raytheon. Certain government and private craft have volunteered as target vessels and the local pilots' associations are also cooperating in the experiment.

Among the problems to be worked out are reliable radiotelephone communications between ships and the information center at Fort Wadsworth and a sure means of identifying craft from the radar scope.

MIT Promotions

FACULTY promotions at Massachusetts Institute of Technology which will become effective July 1 were recently announced. Those taking place in the department of electrical engineering are as follows:

William H. Radford of Watertown is raised to the rank of full professor; Donald P. Campbell and Robert M. Fano of Brookline are promoted from assistant to associate professors; F. Ralph Kotter, Jordan J. Baruch, Elery F. Buckley and James M. Ham of Cambridge, Ronald E. Scott of West Newton and Albert B. Van Rennes of Weston become assistant profess-Maurice Boisvert of Winthrop, Edward J. Craig of West Newton, and Cedric F. O'Donnell, Arthur L. Pike, William M. Siebert, Kenneth N. Stevens and Robert E. Turkington of Cambridge are advanced to the rank of instructor.

International Standardization Makes Progress

CONTINUING progress is being made in the direction of international standardization of electric and magnetic magnitudes and units under the auspices of Technical Committee No. 24 of the International Electro-technical Commission. Existing differences of definitions in different countries are being gradually eliminated. This, it is anticipated, will be of substan-

ELECTRONIC MOTOR and WELDER CONTROLS

Just Out

Here is practical help for those who must select, install, or service electronic controls for motors and resistance welders. The book explains circuit operation of tube-operated equipment found in the two major groups of motor and resistance-welding control. Each chapter starts with an introduction to a complete electronic equipment and then splits it into its component circuits, pinpointing each operation with simple descriptions and diagrams. By George M. Chute, 348 pp., 187 illus., \$6.50

ANTENNAS

2. A new book that thoroughly covers basic antenna theory, methods of antenna measure-

covers basic antenna theory, methods of antenna measurements, and their engineering applications to antennas for all frequency ranges. Treats the significant topics of point sources—power, field, and plass patterns—mutual impedances of antennas—end fire and bload-side arrays—linear, loop, slot, helical and other types of antennas . . stressing the practical aspects of antenna problems throughout. Includes recent advances in the field, and scores of diagrams, equations, tables, and formulas supplement subjects discussed. By John D. Kraus. 533 pages, 368 illustrations, \$8.00

TRANSMISSION LINES AND NETWORKS

3. An up-to-the-minute guide book of information on the theory and applications of transmission lines and four-terminal networks. Discusses transmission lines from an analysis of a c steady-state lines with no reflection, to special considerations for radio frequency lines, telephone, and telegraph lines, and power transmission lines. Foints up useful material on four-terminal networks, and the application of theory to attenuators, impedance-matching networks, and filters. Provides present-day transmission line charts and their use. By Walter C. Johnson. 367 pages, illustrations, \$5.00

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4. Here is a practical manual that gives the various mathematical methods and physical mechanisms which have been developed for use in automatic computation. Familiarizes you with the general character of computing machines outlines the arithmetic techniques employed in their operation and describes basic circuits used to perform various computing machine functions. Describes computers to provide examples of the integration of techniques and components into complete systems, and discusses switches and gates, arithmetic systems, data conversion equipment, etc. Edited by W. W. Stiffer, Jr., E.R.A., Inc. 440 pages, 90 illus., \$6.50

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tial benefit to electrical engineering throughout the world.

One of the overall objectives of Technical Committee No. 24 is to formulate for general acceptance the MKS (meter-kilogram-second) system, sometimes called the Giorgi System. The MKS system has the advantage that its use results in units of current, voltage, and resistance which are respectively the ampere, the volt, and the ohm. This makes it more suitable for engineering work than the CGS (centimeter-gram-second) system, which is in widespread use. It is recognized, however, that the CGS System has certain advantages for physicists. There is no thought of replacing one by the other, but simply of providing a second system that will be particularly helpful to engineers.

Prior to World War II, meetings of Technical Committee No. 24 were held on an average of about once every two years. Due to the war and later the death of the committee's chairman, Professor Kennelly of MIT, no meetings were held from 1938 until last summer when the Committee met in Paris on July 17 and 18.

Questions raised by delegates at the Paris meeting were also proposed for consideration at the Committee's next meeting. These included (1) the most satisfactory name to be assigned to the MKS System; (2) the name to be assigned to the unit of induction; and (3) possible merger of Technical Committees No. 24 and 25. Opinions are being sought from all of the National Committees concerning these matters. This work in the United States is being done through the Sectional Committee on Electric and Magnetic Magnitudes and Units, C61, under the procedure of the American Standards Association, 70 E. 45th St., New York, N. Y.

Cooperative Research Program Announced

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Jerry Golten Co. 2750 W. North Ave. Sbicago 22, III. Martin P. Andrews Mott Road Fayetteville, N. Y. Perimuth-Colman & Associates 1335 South Flower Los Angeles, Cal. Jose Luis Pontet Cardoba 1472 Buenos Aires ian economy was recently announced.

Both Philco and the Institute have cooperated for the past 10 years in a program of training engineers, in many forms of electronic development work and in the exchange of technical information. Under the new agreement both will exchange information on research results in a variety of fields. Military and commercial electronics, television and radio are among the subjects of joint interest.

The program also calls for special conferences and seminars, an exchange of visits by MIT and Philco executive and technical personnel, use of scientific libraries and a joint policy to cover inventions and patents.

BUSINESS NEWS

CANNON ELECTRIC Co., is the new name of the Cannon Electric Development Co., Division of Cannon Mfg. Corp., Los Angeles, Calif., manufacturers of plugs and signal equipment.

KEYSTONE CARBON Co., St. Marys, Pa., has recently completed the addition of 20,000 sq ft of floor space for the manufacture of powdered metal parts and negative-temperature-coefficient resistors.

Philco Corp. has purchased three new plants in Bedford, Ind., about 65 miles southwest of Indianapolis. This will add about 175,000 sq ft to the company's overall manufacturing facilities of 4,200,000 sq ft.

RAYTHEON MFG. Co., Newton, Mass., has announced plans for the erection of a new plant in Quincy, Mass., by the Receiving Tube Division of the company. The 86,000-sq ft plant will be devoted exclusively to the manufacture of subminiature-type electronic tubes.

WESTINGHOUSE ELECTRIC CORP. has purchased 75 acres on a corner of the Baltimore Friendship airport to build a plant for the production of automatic computers to direct gun and rocket fire, radar and autopilots for fighter planes and guided missiles, and complete airborne ar-

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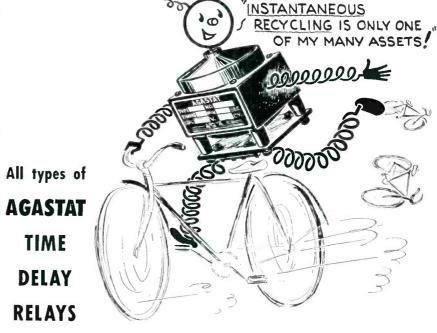
A complete line of photoelectric and electromagnetic detectors, for counting any material or action, can be supplied.

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This new G-E Ozone Lamp emits ultraviolet energy that creates ozone. Ozone chemically destroys or masks many disagreeable odors and substitutes the fresh smell of outdoors.

Extra Sales for Your Product

Among its useful applications are:

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its value and appeal.

As the Ozone Lamp has a mild germicidal effect, it may be used in small sanitary storage cabinets.

To use the ozone lamp a ballast is required. For complete information on this amazing new lamp, write or phone your nearest General Electric Lamp office.

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mament systems. The new 400,-000-sq ft plant is expected to be completed by autumn.

ZENITH RADIO CORP. has begun excavation on its new 453,000-sq ft plant at 1500 N. Kostner Ave., Chicago, Ill. The plant will be used for manufacturing components and the assembling and testing of electronic equipment for the military services.

NATIONAL ELECTRONIC MFG. CORP., Long Island City, N. Y., manufacturers of quartz crystals, crystal holders and other electronic parts during World War II, has resumed production of these items to meet current military demand.

THE WORKSHOP ASSOCIATES, INC., Needham, Mass., recently became a wholly-owned subsidiary of The Gabriel Co., Cleveland, Ohio.

LISTON-FOLB, Div. of Atlas Coil Winders, has now incorporated as Liston-Folb Instrument Co., Inc., and has moved into new quarters at 20 Beckley Ave., Stamford, Conn., thus expanding facilities for production of d-c breaker amplifiers and other special electronic equipment.

JERROLD ELECTRONICS CORP., manufacturer of multiple tv antenna systems, has expanded production and laboratory facilities by moving to new plant space at 26th and Dickinson Sts., Philadelphia, Pa.

MCCORMICK SELPH ASSOCIATES, designers and manufacturers of glass-to-metal seals for guided-mis-



sile and other specialized applications, have moved from Belmont, Calif., to new quarters in Palo Alto, Calif.

Andersen Laboratories, Inc., West Hartford, Conn., has taken over the business and personnel of

The tiny 4-watt size makes it easy to

design into your product - adding to

2 KW VACUUM TUBE BOMBARDER OR **INDUCTION HEATING UNIT**



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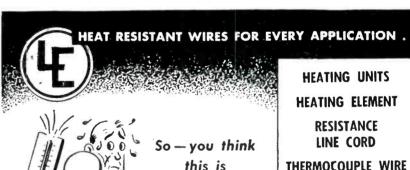
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Today's modern equipment demands a compactness not believed practical a few years ago.

To meet today's demands for a condenser the size of a penny, JOHNSON designed these miniature air variables — the smallest air variables ever built!

Here are **real** space savers, both in front of and behind the panel. Ceramic plates measure only \%" wide by \%" high!

In spite of their extremely small size, these diminutive condensers, accurately produced to close tolerances, are as precisely and carefully built as a fine watch . . . are completely dependable.

Three types are available: The customary Single Section, a Differential and a Butterfly.

These tiny condensers are unequaled for TV and FM applications, laboratory and test equipment, military and other communications, and in a multitude of VHF and UHF uses where compactness and peak performance are demanded.

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Cat. No.	Max.	Min.		
5M11	5.0	1.5		
9M11	. 8.7	1.8		
15M11	. 14.2	2.3		
20M11	.19.6	2.7		

DIFFERENTIAL

Cat. No.	Max.	Min
6MA11	5.0	1.5
9MA11	. 8.7	1.8
15MA11	14.2	2.3
19MA11	19.6	2.7

BUTTERFLY

	mmf		
Cat. No.	Max.	Min.	
3 M B 1 1	. 3.1	1.5	
5MB11	. 5.1	1.8	
9MB11	. 8.0	2.2	
11 MB11	.10.8	2.7	

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- * Single hole mounting bushing threaded 1/4-32, with flats to prevent turning.
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JOHNSON miniature air variables are available in production quantities with special features such as: 1. Locking bearing, illustrated at left, 2. With 180° stop, 3. Various shaft extensions, 4. .0135" spacing offering capacities up to 30 mmfd., 5. High torque.

You are invited to submit your engineer's specifications and quantity desired. We will be happy to quote prices.

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the Andersen-Shaw Laboratories and is continuing research and development work in the ultrasonic field. In addition the new company is manufacturing ultrasonic delay lines and ultrasonic generators.

PERSONNEL

W. B. Whalley, engineering specialist for Sylvania Electric Products Inc., Bayside, N. Y., was recently appointed adjunct professor in electrical engineering for the Polytechnic Institute of Brooklyn.





W. B. Whalley

L. C. Kunz

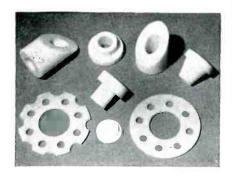
Louis C. Kunz was recently promoted from section engineer on c-r tubes at GE's Electronics Park to product manager for c-r tubes in the company's tube division, with headquarters at Schenectady, N. Y.

EMANUEL R. PIORE, formerly deputy for natural sciences at the Office of Naval Research, Washington, D. C., has been appointed deputy chief of research and chief scientist of the ONR.

FREDERICK N. JACOB, formerly chief engineer of Aladdin Radio Industries, Inc., has joined Belmont Radio Corp., Chicago, Ill., as products manager for infrared equipment.

EUGENE F. GRANT, former chief of the Computer Branch at Air Force Cambridge Research Laboratories, has been appointed to the post of coordinator of electronics research and development for the W. L. Maxson Corp., New York, N. Y.

GEORGE H. PHELPS, formerly associated with RCA Victor, was recently appointed chief engineer of Hammarlund Mfg. Co., Inc., New York City, manufacturers of commercial receivers, selective calling



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• It's power factor is less than 0.05% over the entire spectrum measured to date, 60 cycles to 30,000 megacycles. Its volume resistivity is greater than 1015 ohm-cm. even after prolonged soaking in water. The surface resistivity of the plastic is quite high and drops to only 1013 ohms at 100% relative humidity. "Teflon" also shows good arc resistance. On exposure to an arc, the material is decomposed to a vapor which leaves no carbonized path regardless of time of exposure to the arc. Short time dielectric strengths are high. These values range from 1000 to 2000 volts per mil., depending on thickness. "Teflon" is almost as good in this respect at 200°C (392°F) as at room temperature.

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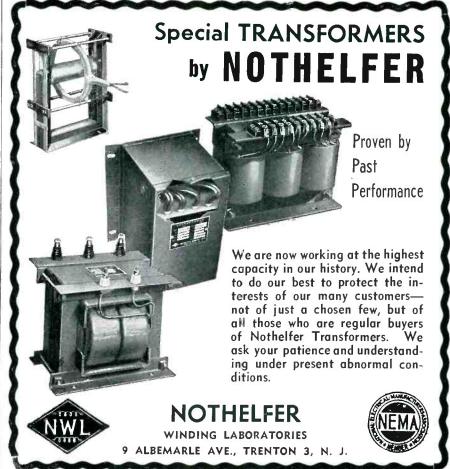


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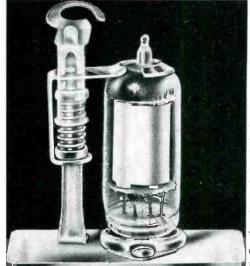
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POSITIVE PROTECTION AGAINST LATERAL AND VERTICAL SHOCK!

The New Birtcher Type 2 Tube Clamp holds miniature tubes in their sockets under the most demanding conditions of vibration, impact and climate. Made of stainless steel and weighing less than ½ ounce, this New clamp for miniature tubes is easy to apply, sure in effect. The base is keyed to the chassis by a single machine screw or rivet...saving time in assembly and preventing rotation. There are no separate parts to drop or lose during assembly or during use. Birtcher Tube Clamp Type 2 is

all one piece and requires no welding, brazing or soldering at any point.

If you use miniature tubes, protect them against lateral and vertical shock with the Birtcher Tube Clamp (Type 2). Write for sample and literature.

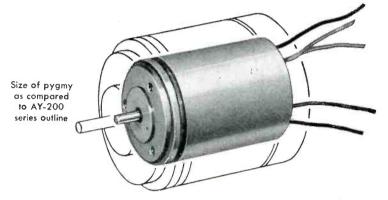
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PYGMY **SYNCHROS**



Eclipse-Pioneer has added a tiny new member to its great family of famous Autosyn* synchros. It's the new AY-500 series, a precision-built pygmy weighing only 13/4 oz. while scaling only 1.278" long and .937" in diameter (the same diameter, incidentally, as a twenty-five cent piece). Its accuracy and dependent ability are assured, thanks to Eclipse-Pioneer's 17 years of experience and leadership in the development of high precision synchros for aircraft, marine and industrial applications. For more detailed information on the AY-500 and other E-P Autosyns, such as the remarkably accurate AY-200 series (guaranteed accuracy to within 15 minutes on all production units), please write direct to Eclipse-Pioneer, Teterboro, N. J.

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	One AY-201-3 Driving		One AY-500-3 Driving	
	One AY-500-3	Two AY-500-3	One AY-500-3	
	Control Transformer	Control Transformer	Control Transformer	
INPUT	26 welte einele abeee	26 volto sinale abase	20 valta simala ubasa	
Voltage	26-volts, single-phase	26-volts, single-phase	26-volts, single-phase	
Frequency	400 cycles	400 cycles	400 cycles	
Current	88 milliamperes	110 milliamperes	55 milliamperes	
Power	0.8 watts	1.2 watts	0.9 watts	
Impedance	105+j280 ohms	100+j220 ohms	290+j370 ohms	
OUTPUT Voltage Max. (rotor output) Voltage at null Sensitivity Voltage phase shift System accuracy (max. possible	17.9 volts	16.2 volts	14.1 volts	
	40 millivolts	40 millivolts	40 millivolts	
	310 millivolts/degree	280 millivolts/degree	245 millivolts/degree	
	23 degrees	26 degrees	44 degrees	
spread)	0.6 degrees	0.6 degrees	0.75 degrees	

Other E-P precision components for servo mechanism and computing equipment: Servo motors and systems • rate generators • gyros • stabilization equipment • turbine power supplies • remote indicatingtransmitting systems and special purpose electron tubes.

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ECLIPSE-PIONEER DIVISION of

TETERBORO, NEW JERSEY



and remote control apparatus, variable air capacitors and radar components.

JOSEPH A. IDANK has joined the staff of Parco Design Corp., New York, N. Y., as chief engineer of the Special Devices Division where he will direct research, development and engineering on precision mechanical and electronic devices and instruments for use on highly specialized industrial and Armed Services projects.

H. B. FANCHER, with General Electric Co. since 1936, has been promoted to division engineer of commercial products for the company's Commercial Equipment Division, Syracuse, N. Y.





H. B. Fancher

R. H. Dorf

RICHARD H. DORF, former audio and ty systems consultant, has joined the staff of Brach Mfg. Div. of General Bronze Corp. as electronics project engineer. He will take charge of the company's government electronics projects.

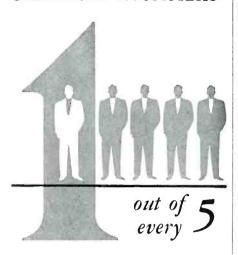
JOHN J. FARRELL, formerly engineer in charge of commercial products for General Electric Co.'s transmitter division, has been appointed assistant manager of engineering for the company's commercial equipment division.

REX WILSON, formerly chief engineer of Tele-Vogue, Inc., Chicago, Ill., has been elected vice-president in charge of engineering for the company.

W. G. HENKE, formerly test equipment design engineer of Admiral Corp., has joined Tele-Vogue, Inc., as coordinator of engineering.

E. B. STEINBERG, previously affiliated with Remington Rand Inc., is now associated with The Reflectone

SELENIUM RECTIFIERS



Here at International Rectifier Corporation, we take our engineering seriously. As a result, one out of every 5 of our employees is a carefully trained, graduate engineer. That means intelligent control of quality from design to finished product. That also means, when you specify International R. C. Rectifiers, you get peak performance and long life.

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2. Water and pressure tight sealing by use of individual seal rings between each contact and insert and between insert and shell.

- 3. Method for adjusting orientation of cable entry fitting.
- 4. Central double-lead lock screw (with bale) for positive engagement and

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disengagement, assurance against accidental disconnection, grounding means for metal shell and shell cap, reduction of space as required by other types of metal shells.

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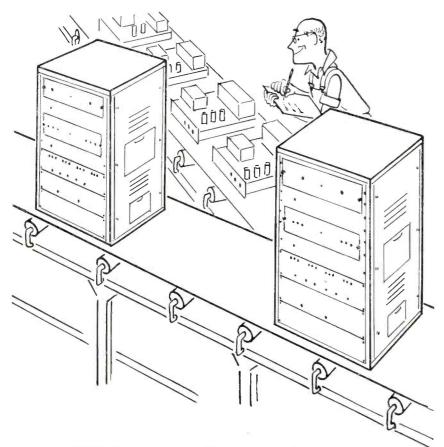
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Corp., Myano Lane, Stamford, Conn., as engineer in charge of its science division.

HAROLD W. GIESECKE, recently promoted to assistant to the general manager of Bendix Radio Division of Bendix Aviation Corp., had been previously associated with Westinghouse Electric Corp. Pittsburgh, Pa.

ADOLPH H. ROSENTHAL, inventor of the Skiatron dark-trace c-r tube, has been elected vice-president and director of research and development of Freed Radio Corp., New York, N. Y.





A. H. Rosenthal

J. D. Schantz

JOSEPH D. SCHANTZ has been promoted from assistant chief to chief engineer of Bell Aircraft Corp., Buffalo, N. Y. As a member of the Radio Technical Planning Board No. Six, he had helped to develop the code of tv standards in use today.

JOSHUA SIEGER of White Plains, N. Y., a member of the technical committees of RTMA and NTSC, will now act as an independent technical consultant in the fields of radar, communications and television.

ROBERT M. ULMER of Cleveland, Ohio, has joined Horizons Inc., of the same city, as a laboratory assistant.

CURTIS B. PLUMMER, formerly chief engineer, has been named head of the new Broadcast Bureau of the Federal Communications Commission.

LEON HILLMAN, previously with New York University's Research Division, was recently appointed head of the engineering department of Production Research Corp.,

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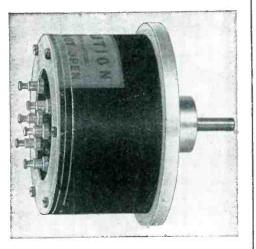
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Thornwood, N. Y., specialists in research, development and production of electronic equipment for the Armed Forces.

DEVEREAUX MARTIN, formerly chief engineer at Radio Receptor Co., Inc., Brooklyn, N. Y., has been named assistant to the president, to work on special assignments, particularly in the field of new-product exploration.

JOSEPH B. HERSH, with the Utility Electronics Corp., Newark, N. J., for the past three years, has been named executive vice-president and general manager of the company. He was formerly chief engineer with the David Bogen Co., Inc.

L. H. JUNKEN has been promoted from designing engineer to division engineer of engineering services for General Electric Co., commercial equipment division, at Syracuse, N. Y.





L. H. Junken

P. L. Spencer

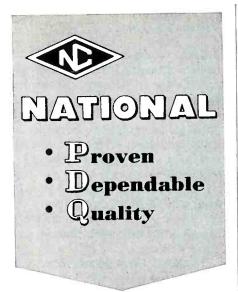
PERCY L. SPENCER, vice-president in charge of the Power Tube Division of the Raytheon Mfg. Co., Waltham, Mass., was recently awarded priority of invention of basic magnetron strapping by the U.S. Patent Office.

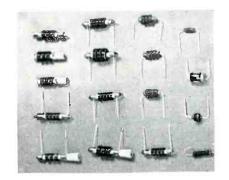
ALFRED C. HAEMER, JR., a member of the National Research and Development Board, has been named head of the field engineering department of General Precision Laboratory, Inc., Pleasantville, N. Y.

GEORGE O. SMITH, author of instruction manuals covering the technical use of sonar, radar and other specialized electronic devices during World War II, was recently appointed manager of components engineering at Emerson Radio and Phonograph Corp., New York, N. Y.

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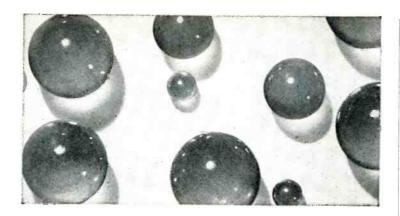




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NEW BOOKS

(Continued from page 150)

drogen phosphate (KDP) and barium titanate. While the treatment is always attentive to theory, the experimenter will find much of value in the substantial amount of practical information on measurement procedure, orientation of cuts and values of physical constants. Of special interest to those who follow theoretical development in this field, will be Dr. Mason's theory on ferroelectricity, more completely presented in this work than in any other writing.

The final three chapters of the book discuss the application and production of ultrasonic waves to measure the properties of gases, liquids and solids. Here too the experimenter will find much of interest in the procedures described, the sketches of equipment and the many graphs and tables of measured data.

It need hardly be said that this work, written by one of the foremost researchers of a laboratory distinguished, among other things, for its contributions to this field is strongly recommended to all who have serious interest in the properties of crystals.—J. F. MCALLISTER, General Electric Co., Syracuse, New York.

Vacuum-Tube Voltmeters

BY JOHN F. RIDER. Second edition, published by John F. Rider Publisher, Inc., New York, 422 pages, \$4.50.

THE IMPORTANCE of the vacuumtube voltmeter has been stressed repeatedly. Many engineers, however, overlook some of the less obvious jobs that can be done with vtvm's. It is hard to imagine a person involved in laboratory work who could not discover at least five new ideas from reading Rider's book.

This second edition brings the first (1941) edition up to date. The applications section alone will be worth the price of the book to many. It gives easy-to-follow instructions for measuring such things as inductance, resistance, capacitance, r-f, a-f and d-c voltages and currents with a vtvm. A completely new chapter deals with d-c and r-f probes of all types.

One section is devoted to a comparison of the more than 40 com-



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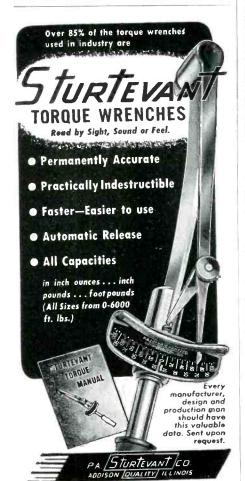
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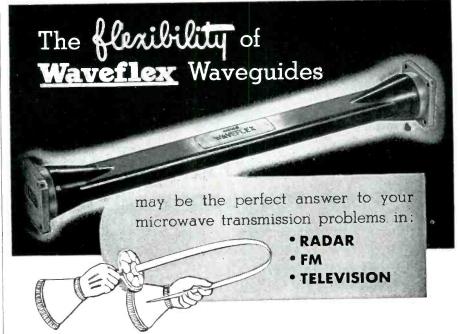
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Transformers

By F. C. CONNELLY. Pitman Publishing Corp., New York, 1950, 490 pages, \$7.50.

IN RESEARCH work involving electrical and electronic circuitry, the need often arises for transformers having special characteristics. Quite often the circuit specialist finds himself stymied by the nonexistence of an appropriate transformer for his needs. In some of the more elaborate setups, special shops and trained personnel are provided for such cases. often, the engineer himself must rewind or start from scratch and prepare his own.

This British-authored book is designed as a guide in such cases. Carefully read, it can furnish the general engineer with transformer know-how and allow him to proceed logically and methodically in designing a special transformer for a specific application. Well-planned step-by-step procedures are suggested.

Basic principles are supplemented by separate sections on even the most advanced details of transformer work. Included are separate sections on multi-phase. power, high-voltage, vibrator, instrument, audio and output transformers, as well as autotransformers. A short section in the back of the book is devoted to special television types. Among the theoreti-

Detroit 26

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cal discussions are chapters on magnetic leakage, core excitation, transients, losses, efficiency and regulation.

The introductory material is quite easy to understand. The analogies between the electrical circuit and the magnetic circuit are clearly emphasized and helpful in creating an understanding of the latter. The book is highly recommended for the beginner or the expert.—J.F.

Semi-Conductors

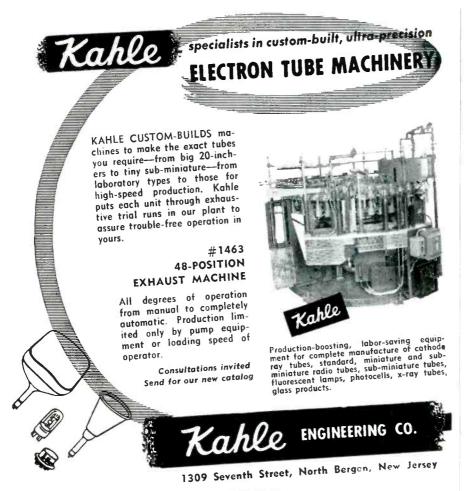
By D. A. Wright. Methuen's Monographs, John Wiley & Sons, Inc., New York, 1951, 130 pages, \$1.75.

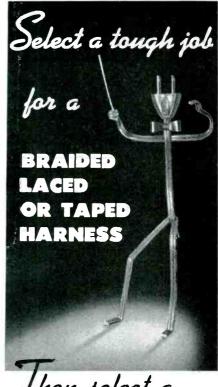
THIS condensed, elementary treatof basic semi-conductor ment theory should be of considerable value as a means of review for the physicist or as an introduction to the subject for the student. Engineers could particularly benefit because its logical sequence of presentation of concepts would give them a basic understanding of the subject and permit more intelligent use of the more rigorous and fundamental texts.—J. P. JORDAN, Electronics Laboratory, General Electric Co., Syracuse, N. Y.

Receiver Circuitry and Operation

By A. A. GHIRARDI AND J. R. JOHNSON. Rinehart Books, Inc., New York, 1951, 669 pages, \$6.00.

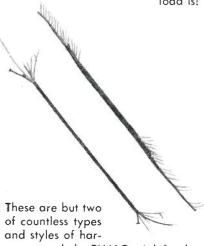
HERE is basic coverage of the circuits most often encountered in modern television, a-m radio and f-m radio receivers, written for the technician and student who seeks to bring his education up to date. The treatment is nonmathematical and has a minimum of complicated theoretical analysis of circuit operation. Even the number of graphs is held to an absolute minimum. A knowledge of radio and electrical fundamentals, including vacuumtube theory, is assumed to have been obtained elsewhere; practical servicing and troubleshooting procedures are likewise absent but are scheduled for a second volume now in preparation. Chapters dealing with construction and operation features of components and acces-





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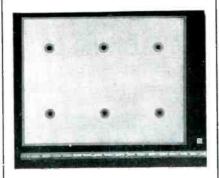
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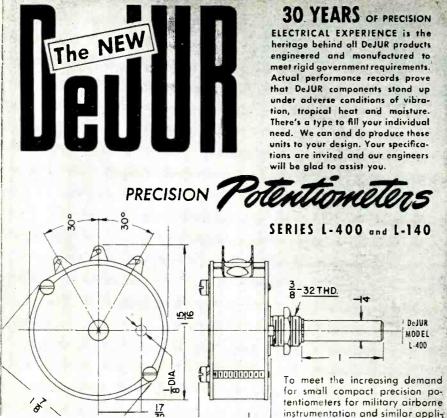
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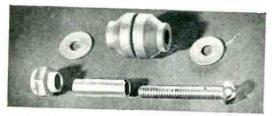
sories include: Loudspeakers: Receiving Antenna Systems; Home Recorders; Phono Pickups and Record Players; Automatic Record Changers; Mechanical Construction of Receivers.

Explanations in this book are clear and concise throughout, and questions at the ends of chapters provide a thorough test of mastery of the contents. Equally deserving of commendation are the excellent drawings, which show that the artist understood what he was doing. Long waited for, the book goes far toward filling the demand for up-to-date information in a style comparable to Ghirardi's stillselling 1931 edition of Radio Physics Course and his almost equally out-of-date Modern Radio

Advances in Electronics

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(continued)

vide a clear picture of absorption phenomena, and an analysis of theoretical and practical resolution capabilities is presented.

The second chapter contains information compiled by R. R. Warnecke, P. R. Guenard, M. Chodorow, and E. L. Ginzton. The first two are affiliates of the Laboratories de Recherches de la Compagnie Générale de T.S.F., Paris, France; the latter two are on the staff of Stanford University, Stanford, California. This chapter discusses velocity-modulated tubes, beginning with the basic concepts that led to the development of the klystron, and working into even the most detailed aspects of klystron operation in low-noise, low-power and power klystron amplifiers.

Electronic theory of the plane magnetron is discussed by L. Brillouin, of International Business Machines Corporation of York, in the third chapter. Here again the discussion starts off with basic concepts and evolves into a detailed description from beginning to end in the field of plane magnetrons. Cylindrical magnetron history and theory are presented by L. Brillouin and F. Bloch of International Rusiness Machines Corp. and Stanford University respectively. Through the engineer's medium of curves and equations, the subject is covered quite concisely.

John E. White, of the National Bureau of Standards at the time of writing but now with General Electric Co., Schenectady, describes tube miniaturization. Physical and electrical problems peculiar to the subminiature tube field are presented. The next chapter serves to supplement White's chapter by giving details on subminiaturization techniques. In this chapter Gustave Shapiro, of the National Bureau of Standards, presents a general picture of the methods and problems of constructing electronic equipment using subminiature tubes and components. Much of the subminiature information presented has been set down elsewhere, in various manufacturer's catalogs, but the organization of this summary is quite good, and the subject made interesting.

H. F. Mayer, of the School of



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Electrical Engineering, Cornell University, Ithaca, New York, writes on principles of pulse code modulation, with emphasis on noise, power and bandwidth. Next is presented a summary of modern methods of network synthesis by E. A. Guillemin of Massachusetts Institute of Technology. Theoretical and practical methods are covered, with typical examples furnished as illustrations.

The final chapter, prepared by Meyer Leifer and William F. Schreiber of the Sylvania Electric Products Inc., Bayside, New York, is an excellent discussion of communication theory. The authors have written from a standard outline on this subject, but their individual explanations are helpfully clear and easy to follow. They begin conventionally with summaries of the contributions of Hartley, Gabor, Tuller, Goldman and Shannon, and carry the theoretical analysis straight through to application in television and related fields. Here again sections of the analysis will be recognized as having been taken directly from the literature, but the organization increases their individual value many times.

The book is conveniently indexed, and a moderate number of pertinent references are included. The engineer who wishes to increase his knowledge beyond his particular project would do well to acquire a copy of *Advances in Electronics*, Volume Three.—J.D.F.

THUMBNAIL REVIEWS

THE ENGINEERING METHOD. By J. C. Fish. Stanford University Press, Stanford, Calif., 1950, 186 pages, \$3.00. Philosophy and practice of reaching engineering decisions as to which method, which equipment and which material to use in solving a particular problem. Comprises separate publication of chapters 9, 10 and 11 written by this author for "The Engineering Profession" (Hoover and Fish, Stanford University Press, 2nd edition, 1950).

MATHEMATICAL ENGINEERING ANALYSIS. By Rufus Oldenburger. The Macmillan Co., New York, 1950, 426 pages, \$6.00. Aimed at easing the transition from physical engineering situations to corresponding mathematics. Divided into five major parts, one of which is Electricity and Magnetism; this covers fundamentals, electric and electronic circuits and systems, tubes, transistors,

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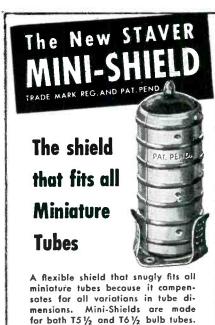
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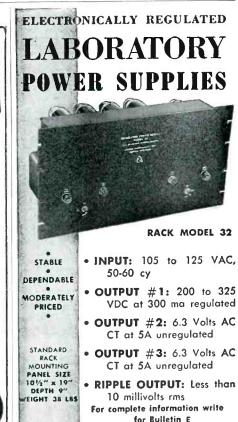
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TELEVISION AND FM ANTENNA GUIDE. By E. M. Noll and M. Mandl. Macmillan Co., New York, 1951, 311 pages, \$5.50. Combination textbook and practical installation guide. First part gives principles of antenna systems, including wave propagation and transmission line theory. Second part, arranged as reference guide, covers antenna site surveys, choice of antenna, erection, installation of transmission line, use of boosters, characteristics of various antenna types, rotating antennas, multipleoutlet antenna systems, and interference reduction.

ALTERNATING-CURRENT ELECTRI-CAL SYSTEMS FOR AIRCRAFT. West-inghouse Electric Corp., P. O. Box 2099, Pittsburgh, 70 pages loose-leaf, \$1.50. Latest design requirements, maintenance hints, and data on various aspects of a-c power supply systems for large aircraft.

ADVANCES IN RADIOCHEMISTRY. By E. Broda. Cambridge Monograph on Physics, Cambridge University Press, New York, 1950, 152 pages, \$2.75. Production of radioelements, chemical properties, and aspects of transmutation relevant to nuclear synthesis. Final chapter covers new developments in technique of measuring radioactivity. radioactivity.

NUCLEAR SCIENCE ABSTRACTS, Cumulative Index to Vol. 1, 2, 3 and 4. U. S. Atomic Energy Commission, Technical Information Service, Oak Ridge, Tenn., 1951, 241 pages, \$.60. Cumulative author index, subject index, nuclide index, foreign geographic index, numerical index of AEC reports with indications of availability, list of AEC depository libraries and list of periodicals abstracted.

TELEVISION PROGRAMMING AND PRODUCTION. By Richard Hubbell. Rinehart & Co., New York, 1951, 240 pages, \$4.50. Second edition, revised and enlarged, with technical chapters brought up to date and with much new program material and many new illustrations.

UNDERSTANDING RADIO. By H. M. Watson, H. E. Welch and G. S. Eby. McGraw-Hill Book Co., New York, 1951, Second Edition, 716 pages, \$5.50. Revision of fundamental text for students with little or no background in electricity or science. science.

SOUND SLIDEFILM GUIDE. Operadio Mfg. Co., St. Charles, Illinois, 1951, 62 pages, \$1.00. Over 1,000 available sound sildefilms are listed, each with data on source, availability (free loan, rental rate or sale price), running time and nature of film. Includes 40 sound slidefilms giving technical training in electronics.

FOURIER TRANSFORMS. Ian N. Sneddon, University of Glasgow. International Series in Pure and Applied Mathematics, McGraw-Hill Book Co., New York, 1951, 542 pages, \$10.00. Theory of Fourier transforms and related topics, presented in a form suitable for students and research workers interested in boundary-value problems of physics and engineering. Seven of the ten chapters illustrate the use of the theory in solving problems of vibration, heat conduction, slowing of neutrons, hydrodynamics, atomic and nuclear physics, and elasticity.

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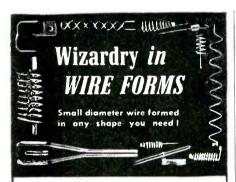
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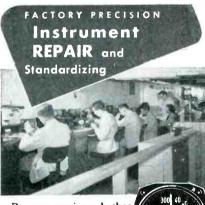
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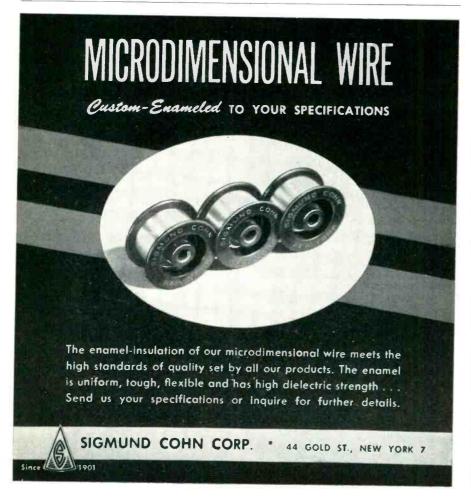
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BACKTALK

(continued from page 152)

questions raised in the various symposia of the recent IRE convention in New York. It appears that a number of persons are interested in certain branches of medical electronics which are outside the scope of existing professional groups. However, the organization of such a group is quite a task at the best and it may be to our advantage to organize it within the IRE.

I have been assured that the IRE is interested and will do what it can to help. The task at hand now is to locate those persons who wish to become members of this group. The location of these people is complicated by the fact that they are scattered geographically, and in addition the papers they have published on medical-electronic subjects are scattered through all types of technical and medical publications. Therefore I am turning to ELECTRONICS to help locate them through this letter.

In other words, if any one is interested in becoming a member of the medical electronics group of the IRE, when organized, it would be appreciated if he would send me his name, address and state whether or not he is now a member of the IRE. Those interested will then be placed on the mailing list and notified as to the progress of the organization.

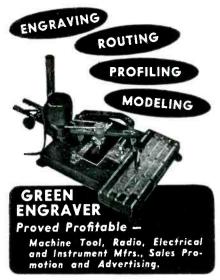
L. H. MONTGOMERY

Dept. of Anatomy
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Too Big

DEAR SIRS:

I HAVE JUST finished reading your editorial in May 1951 Crosstalk on the recent six-ring circus otherwise known as the IRE Convention and I want to take this opportunity to assure you that there are many others who agree with your sentiments. It was all too obvious to many that the attempt to funnel the large number of people present into six simultaneous sessions was disastrous, and that those who wished to pick up an early paper in one session and then transfer to another found themselves just as you mentioned-"out of luck." You may be sure that your final com-



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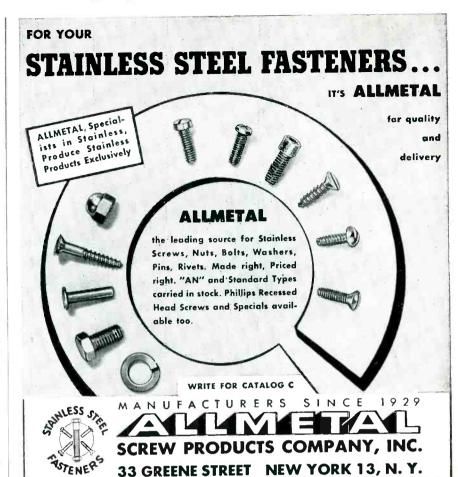
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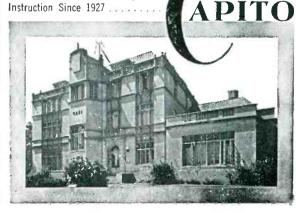
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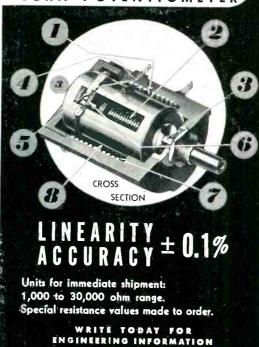


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ment in regard to the use of "a sharp and heavy axe" has found a responsive chord in me.

I have heard numerous expressions to the effect that it would have been much better to have had a major reduction in the number of papers and a major improvement in their technical excellence.

Since this is the first time I have ever written to an editor of any magazine, I may as well comment further on the sound editorials which now appear over the signature of McGraw-Hill Publishing Co., Inc. The sound advice on economic questions offered so many times in the past would seem sufficient reason for placing the President and the members of our Congress on your mailing list.

Ellsworth D. Cook General Engineering Laboratory General Electric Co. Schenectady, New York

And, It Works!

DEAR SIRS.

ALTHOUGH I do not subscribe to ELECTRONICS, I always try to read every issue. For quite some time I have been doing a lot of experimental work with neon relaxation oscillators for use in electronic organs and I won't hesitate one moment to say these simple little oscillators have almost driven me to the point of throwing everything into the junk box and quit-

I have tried and tested all circuits that have appeared lately in other magazines, but all these articles and circuits dwell on theory as to what these circuits should do. But they never work out in practice. Every article and writer has had a different theory on the way the neon tubes are to be synchronized with the master oscillators. The main difficulty was, of course, with frequency stability.

Then along comes a swell article on "Gas-Diode Electronic Organ" by Robert M. Strassner in your January 1951 issue. After all the hard work up to now I decided to give the circuits shown in his article a try. What a surprise. For the first time in months I had something put together that really worked.

My research into the subject of



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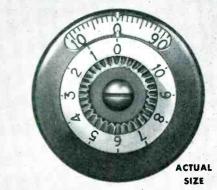
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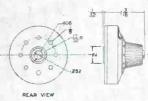
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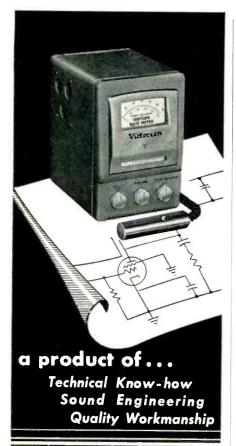
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electronic organs has covered all articles and every circuit I could lay my hands on, but Strassner's way of using an electrostatic coupling between the master oscillators and the neon relaxation oscillators in the octave divider chain is the best I've seen and tried.

As you must already know, this subject of electronic organs has become quite popular in the last year or so, and is given quite some space in other magazines; Richard Dorf, writing these articles, has been snowed under with letters asking for more information on the subject. These articles of his offer nothing concrete, but just enough to get the fellows started.

With what Strassner offered in his first article, why can't we have another by him and let him go a little deeper into the subject of master oscillators and the way the switches are mechanically linked to the playing keys. This subject demands another article giving a complete schematic and other constructional details.

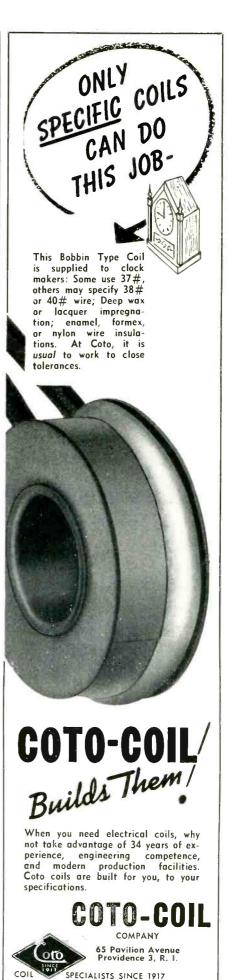
Robert Strassner has kindled the spark in lots of fellows like myself, and now all we need is another good article to keep the fire going. I know that a swell magazine like ELECTRONICS won't let us readers down.

GLEN E. TURNER Fresno, California

Student Answers Briton

DEAR SIRS:

I SHOULD like to comment on the letter by C. H. Banthorpe published in the April, 1951 issue of ELEC-TRONICS. One might almost say I should like to take issue with him, for I feel that his statements are rather dogmatic on the basis of what has apparently been somewhat limited listening to vhf programs. There are no doubt many desirable features of a-m for vhf broadcasting, not the least of these being the use of the same detector for both the medium-frequency band and the vhf band during this changeover period when in most communities reception of both media is necessary to full program coverage. However, I should like to point out some considerations which





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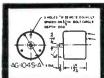
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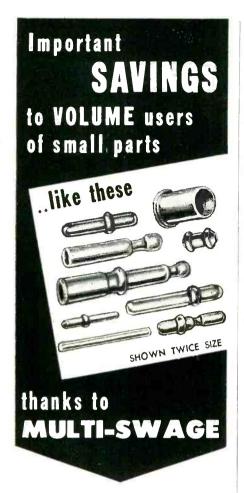


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I believe favor f-m, at the same time inviting criticism of my views by any and all who are qualified.

Banthorpe makes the very good point that comparison of f-m on vhf with a-m on m-f is often misleading. This is quite true. But he then goes on to apparently commit this same misleading comparison in commenting on the relatively complex nature of the vhf f-m receiver compared to the m-f a-m receiver. I cannot see how the method of detection would materially influence the major problems of oscillator stability and low gain per stage. The first appears to be a consequence of the very high frequency at which the local oscillator must operate, and the second to be a consequence of the high intermediate frequency necessary to prevent the local oscillator locking in with the incoming signal. As a matter of fact, the gated beam tube recently developed in this country may make the f-m vhf set less complex than its a-m complement. This tube and very few associated components produces near zero time constant limiting, and is a discriminator that delivers sufficient output to permit doing away with the first audio tube. It also obviates the necessity for an automatic gain control circuit, saving several parts.

Automatic gain control deserves a bit more comment. Fades that reach the lower audio frequencies in their rates of occurrence are not uncommon on vhf. They are particularly noticeable in the presence of nearby aircraft, a problem far from negligible now, and one not likely to decrease. To design a circuit with time constants such that it can deal with these fades with anything approaching the effectiveness of the f-m limiter and which will yet not act to attenuate or distort the lowest modulation frequencies of the a-m signal sounds to me like quite a project, and one not likely to contribute to economy.

And that brings up attenuation of frequencies in the reproduced signal and distortion. In the a-m receiver, if flat response is to be achieved, along with good signalto-noise ratio and adjacent-channel selectivity, the i-f curve must approach the ideal very closely. In

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products.

the f-m set, minor departures from the ideal are cleaned up by the limiter. And since the i-f curve and tuning are so closely linked, it might be well here to point out that if the i-f response were made optimum for adjacent channel selectivity and noise level, the same amount of detuning that causes moderate distortion on f-m (this considered to be a weakness of f-m) would not only cause distortion, but might even tune right off the station completely.

Banthorpe has excited my curiosity about several new English developments. He notes that there is available a simple diode noise limiter of value for broadcast work. This is obviously a considerable improvement over the latest type of noise limiter I have seen for a-m receivers, for this type is capable of limiting noise only to the level of the maximum modulation of the received signal. Since the noise level for good reception of entertainment should be somewhat lower than this, I had always supposed the device to be virtually useless for broadcast reception, even discounting the extremely critical adjustment of the threshold control.

He has also whetted my interest in the BBC tests. I am a little chagrined that the only comment on the method of modulation used in the BBC f-m transmitter is that it is a "simplified design". (This is not Banthorpe's statement, but comes from another publication and is the most description I have so far been fortunate enough to find.) Despite my great faith in the high engineering standards of the BBC, this reticence to comment tends to disturb me. It also appears that Banthorpe has been willing to draw conclusions somewhat before the BBC itself was willing to do so.

Now as to the use of f-m sound with television, in addition to the foregoing we may sight the ease of separation of sight and sound intelligence when different modulation systems are employed for the video and audio portions of the signal. Indeed, the intercarrier technique of beating the sound against the picture carrier to obtain the 4.5 i-f would doubtless be impossible with a-m, and this is one of the



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tools that contributes to low cost of many of the cheaper receivers in this country.

As to overwhelming arguments for 405 lines, suffice it to point out that on the 20-inch tubes now becoming popular for use in even quite small rooms, the lines of such a system would be of the order of 32 inch. The trend here has continually been one of greater definition—a progressive, not a retrogressive trend.

As I commented at the start, I submit these comments with a degree of humility, as befits a college student coming before the learned group that is the ELECTRONICS circulation. If I have at least shown that there might be another side to the discussion, I shall be satis-

> ROBERT F. HOOPER Student Albion College Albion, Michigan



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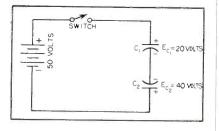
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Electronics Quiz

THIS month's puzzler was submitted by Robert C. Burns of Hicksville, Long Island. For this contribution, Burns will receive our check for \$5.00, as will all other contributors whose problems are accepted and published in this department.

This Month's Problem

Given the following circuit, and assuming a perfect battery and perfect capacitors.



what will be the final capacitor voltages after the switch is closed?

Answer will appear next month.

Last Month's Problem

Last month's problem involved the charging of the capacitor in the circuit diagram of Fig. 1A. It was desired to determine the initial charging current flowing into the capacitor at the time the

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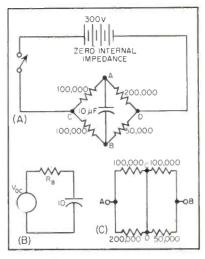


FIG. 1—Solution of last month's quiz

switch was closed, and also to determine the time constant of the circuit.

The answer submitted by the author of the problem is as follows: "By Thevenin's theorem, the circuit can be reduced to that shown in Fig. 1B where V_{ac} is the open-circuit voltage at AB (that is, with capacitor removed) and $R_{\scriptscriptstyle B}$ is the output resistance (the resistance looking back into AB, with all sources of emf replaced by their internal resistance). Clearly, $V_{\circ \circ}$ is 100 volts, and by Fig. 1C, R_B is 100,000 ohms. At the instant of closing the switch the capacitor appears as a shortcircuit, and the initial current is V_{oc}/R_B or 1 ma. The time constant $= CR_{B} = 10 \times 10^{-6} \times 10^{5} = 1$ second.

It is worth noting that the initial current can be found using Kirchhoff's Laws and replacing the capacitor by a short-circuit, but the time constant cannot be easily found in this manner.

Electronic Organ

DEAR SIRS:

SINCE the publication of my article entitled, "Gas-Diode Electronic Organ" (ELECTRONICS, Jan. 1951, p 70), two mistakes in diagrams have been brought to my attention. On page 72, Fig. 2, the lead marked—100 v should be connected to the ground bus. In Fig. 3 the waveforms should show initial discharge peaks at the beginning of each positive and negative half-cycle.

ROBERT M. STRASSNER Monterey Park, Calif.

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ELECTRONIC EQUIPMENT, Vacuum Pumps, etc. VET GLASS SLICING and Cutting machines for Laboratory use ISLER ENGINEERING CO., Inc. 75) So. 13th St., Newark 3, N. J.



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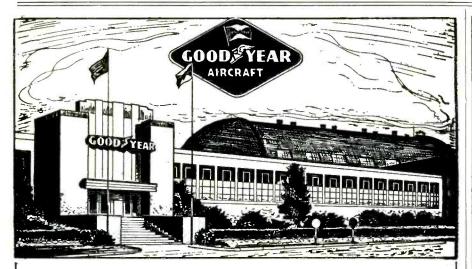
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AN ADVERTISING INCH is measured 1/8 inch vertically on one column, 3 columns—30 inches

NEW ADVERTISEMENTS received at the N. Y. Office, 330 W. 42 St., N. Y. 18, by July 2nd will appear in the August issue, subject to limitation of space available. The publisher cannot accept advertising in the Searchlight Section which lists the names of the manufacturers of resistors, capacitors, rheostats, and potentiometers or other names designed to describe such products.



RESEARCH - DEVELOPMENT - DESIGN

ENGINEERING WITH A FUTURE

The steady growth of several established research and development projects has created a number of exceptional engineering opportunities with a future.

PHYSICISTS—ENGINEERS

Positions are now available in our organization for qualified physicists and engineers with backgrounds in circuit analysis, microwaves, servomechanisms, analog computers, etc. Openings exist at several levels with salaries dependent on education, ability, and experience.

If you are qualified and interested in a position which combines stability and unusual opportunity, write, giving full details to Mr. C. G. Jones, Manager, Salary Personnel.



Akron 15, Ohio

WANTED ELECTRICAL ENGINEERS

The Taylor-Winfield Corporation, 1052 Mahoning Avenue, N.W., Warren, Ohio (a leading manu-facturer of electric resistance welding machines) has positions available for two recently graduated electrical engineers

One position involves work of a design and developmental nature on industrial electronic controls to be used with welding equipment. Basically the duties are those of a project engineer working under supervision to amend theoretical designs, make breadboard analyses and then follow by testing of pilot models. On completion of a successful design proven by tests, complete detailed engineering must be prepared for release to the manufacturing denartment. partment.

The second position involves work of an inspection and test nature associated with field service. Controls, after selection by the Engineering Department are connected to welders and the proper functioning of the welding equipment studied. After physically functioning correctly, actual welding of sample parts is performed. In the more complicated equipment, field service is undertaken to allow the user to rapidly set up equipment and obtain the desired results. This is an excellent opportunity to become familiar with electrical components used in industry, with industrial electronic controls, and with methods of manufacture of many products by visits to plants in which Taylor-Winfield equipment is used.

If interested, please contact the Personnel Department, submitting an analysis of your training and experience in order that a personal interview may be arranged.

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-All Grades-

Men whose education and interest or experience qualifies them for the development or design of airborne or ground radio, radar, or other electronic equipment can work under ideal conditions. We have openings in our small, efficient group (under 20 people) where the work is personalized and individual ability can be recognized and rewarded.

Write or telephone

Chief Engineer

Kellett Aircraft Corporation Central Aircraft Corporation
Central Aircraft, Camden 11, N. J.
Merchantville 8-4800
Founded 1929

REPLIES (Box No.): Address to office nearest you NEW YORK: 330 W. 42nd St. (18) CHICAGO: 520 N. Michigan Ave. (11) SAN FRANCISCO: 68 Post St. (4)

EMPLOYMENT SERVICE

SALARIED PERSONNEL, \$3,000-\$25,000. This confidential service, established 1927, is geared to needs of high grade men who seek a change of connection under conditions assuring, if employed, full protection to present position. Send name and address only for details. Personal consultation invited. Jira Thayer Jennings, Dept. L, 241 Orange St., New Haven, Conn.

(Continued on page 285)

A Reminder

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If you have a Bachelor's or Advanced Degree in Electrical or Mechanical Engineering, Physics, Metallurgy or Physical Chemistry and experience in the Electronics Industry, you can expand your training and experience to the fullest in openings in connection with:

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- .. ELECTRONIC COMPONENTS
- .. TELEVISION, TUBES and ANTENNAS.

Please send resume to: TECHNICAL PERSONNEL ELECTRONICS PARK



OPPORTUNITIES IN RESEARCH AND DEVELOPMENT

IN
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PHYSICS
ELECTRONICS
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MECHANICAL ENGINEERING
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PERMANENT POSITIONS FOR QUALIFIED PERSONNEL EXPERIENCED DESIRABLE BUT NOT ESSENTIAL WORK VITAL TO NATIONAL DEFENSE

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CORNELL AERONAUTICAL LABORATORY, INC.

P.O. BOX 235, BUFFALO 21, NEW YORK



YOU are an engineer or a physicist and

YOU have skills and interests related to diverse armament development projects . . . and

YOU are capable of assuming greater responsibility . . and

YOU prefer the personalized attitudes and policies and the demands for ingenuity and versatility offered by a small progressive company—

WE would like to make your acquaintance





Development Engineers for Electronic Aircraft Armament

ELECTRONIC INSTRUMENTATION • ELECTRONIC COMPONENTS • SERVOMECHANISMS • RADAR • ELECTRONIC PACKAGING • CALIBRATION AND TESTING ENGINEERS FOR PRODUCTION

Job openings range from recent graduates to Engineers with years of experience. Attractive employee benefits include group insurance and pension plans; paid holidays and vacations.

Send complete resume, listing salary requirements and availability, to:

Technical Employment Supervisor, Station 483-D

THE EMERSON ELECTRIC MFG. CO.

8100 Florissant • St. Louis 21, Missouri

LEADERS IN THE ELECTRICAL INDUSTRY SINCE 1890



Unusual opportunities for outstanding and experienced men.

These top positions involve preliminary and production design in advanced military aircraft and special weapons, including guided missiles.

Immediate positions include:

Electronic project engineers
Electronic instrumentation engineers
Radar engineers

Radar engineers Flight test engineers

Stress engineers

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Electro-mechanical designers
Electrical installation designers
Excellent location in Southern Cali-

Excellent location in Southern California, Generous allowance for travel expenses.

Write today for complete information on these essential, long-term positions. Please include resume of your experience & training. Address inquiry to Director of Engineering,

NORTHROP AIRCRAFT, INC.

1009 E. Broadway

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SCIENTISTS! ENGINEERS! DESIGNERS! ORAFTSMEN!

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LONG RANGE PROGRAMS

on

GUIDED MISSILES

AIR-BORNE ELECTRONIC EQUIP'T

ROCKET ENGINES
HELICOPTERS
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Scientific or engineering degree and extensive technical experience required.

INQUIRE NOW!

Manager, Engineering Personnel



BENDIX RADIO DIVISION



TV RECEIVER DESIGN ENGINEERS
—Progressive TV Design Section
has excellent opportunity for highly
experienced men with proven ability. Desirable openings also available for promising young engineers.
We will be pleased to discuss
salaries at all levels.

ELECTRONICS ENGINEERS — At all salary and experience levels.

RESEARCH ON: Antennae, Servomechanisms, Microwave ccts. and other phases of communications and navigation equipment.

PRODUCTION DESIGN OF: Military and commercial communications and navigation equipment.

FIELD ENGINEERS — Supervise installation and maintenance of radio and radar equipment. Factory training will be given. Base salaries from \$4200 to \$6900 per year. 25% bonus for time spent overseas. Traveling and living expenses paid by Bendix. Insurance plan.

TEST AND INSPECTION ENGINEERS — Practical knowledge of radio, radar, or TV manufacturing processes. Good knowledge of radio fundamentals essential. Base salaries from \$3900 to \$5880.

TECHNICAL WRITERS — Knowledge of radar fundamentals or radio required. Work closely with engineers to gather material for instruction and maintenance manuals. Base salaries from \$3400 to \$4300.

LABORATORY TECHNICIANS — Require knowledge of radio fundamentals and skill in use of measuring instruments and laboratory equipment. Previous industrial experience essential. Salaries from \$262 to \$321 per month.

BASE SALARIES FOR ALL POSI-TIONS LISTED ABOVE ARE SUP-PLEMENTED BY UP TO 30% FOR REGULARLY SCHEDULED 48-HOUR WEEK.

Housing is no problem in Baltimore.

Excellent group insurance and family hospitalization plan.

Attractive retirement plan for professional personnel.

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Engineering Personnel Supervisor
Department E
BENDIX RADIO DIVISION of
Bendix Aviation Corporation
Baltimore 4, Maryland
TOwson 2200

ATOMIC ENERGY INSTALLATION

NEEDS

ELECTRONICS ENGINEERS

Two to ten years' experience in research, design, development or test. Patent history desirable but not necessary. A variety of positions open for men, with Bachelor's or advanced degree, qualified in one or more of the following fields:

- RELAYS
- PULSE CIRCUITS
- SERVO-MECHANISMS
- LOW POWER APPLICATION
- TELEMETERING
- UHF TECHNICIANS
- INSTRUMENTATION
- . QUALITY CONTROL

TEST EQUIPMENT RELATING TO THE ABOVE

CAREER DRAFTSMEN

EXPERIENCED CAREER DRAFTSMEN WITH NO COLLEGE DEGREE.

These are PERMANENT POSITIONS with Sandia Corporation in Albuquerque, New Mexico. Sandia Laboratory is operated by Sandia Corporation, a subsidiary of Western Electric Company, under contract with the ATOMIC ENERGY COMMISSION. This laboratory offers good working conditions and liberal employee benefits, including paid vacations, sick leave, and a retirement plan.

Albuquerque, center of a metropolitan area of 150,000, is located in the Rio Grande Valley, one mile above sea level. The "Heart of the Land of Enchantment." Albuquerque lies at the foot of the Sandia Mountains, which rise to 11,000 feet. Climate is sunny, mild and dry the year 'round.

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SANDIA BASE

ALBUQUERQUE, NEW MEXICO

ELECTRONIC ENGINEERS

SENIOR ENGINEERS or PHYSICISTS
Degree and experience in Radar,
Pulse Circuits, Digital or Analogue
Computers, or Servomechanisms
JUNIOR ENGINEERS and recent
graduate in EE or Physics.

ELECTRONIC ENGINEERING COMPANY OF CALIFORNIA

180 S. Alvarado St. Los Angeles 4, Calif.

ELECTRONIC ENGINEER

At least three (3) years post-college experience in development, DC amplifier, digital computers, pulse and servo design. Established Company, New York City.

P-9662, Electronics 330 W. 42 St., New York 18, N. Y.

WANTED

Research and Development Engineers and Physicists with educational background in mechanical, electrical or electronic engineering, physics or engineering physics for openings in plant and laboratory instrumentation, physical measurements, geophysics, and industrial electronics. Prefer persons with two to four years experience in experimental research design and development of instruments, intricate mechanisms, electronic apparatus, optical equipment, servo-mechanisms or allied fields. Positions are of immediate and permanent importance to our operations.

Write Personnel Director

Research and Development Department

PHILLIPS PETROLEUM COMPANY
Bartlesville, Oklahoma

ELECTRONIC ENGINEERS

who have at least 5 years of good experience in Nucleonics, Radar, Guided Missiles, Computors or Allied fields are desired immediately for development work on new instrumentation. These positions give opportunity to exercise judgment and initiative. Ours is a growing concern with progressive ideas. Those associated with this company have real opportunities for advancement. We will be pleased to hear from qualified men. Please send resume to Personnel Director and phone for appointment.

Tracerlab 130 High Street, Boston, Mass.

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TUBE ENGINEERS

PRODUCTION & DEVELOPMENT With transmitting & rectifier tube experience. Write direct stating experience, salary requirements.

TAYLOR TUBES, INC. 2312 W. Wabansia Ave., Chicago 47, III.

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Experienced in the development of electronic storage devices. Excellent opportunity with progressive, established Eastern manufacturer. Moving expenses paid. Interview arranged upon receipt of detailed expenses. Interview uncarding trailed resume.

P.9913, Electronics

330 W. 42 St., New York 18, N. Y.

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GENERAL MOTORS CORPORATION

PRECISION INSTRUMENT PLANT

Positions now available for highest caliber personnel in the field of airborne autamatic electro-mechanical control equip-

MECHANICAL DESIGN ENGINEERS ELECTRONIC ENGINEERS SERVO ENGINEERS JUNIOR ENGINEERS

New and expanding division of an established firm with 20 years of successful experience in the instrument field. Work involved deals with the manufacture and development of highly complex equipment of the most advanced type.

Write or Apply

AC Spark Plug Division GENERAL MOTORS CORPORATION

1925 E. Kenilworth Place Milwaukee 2, Wisconsin

Electrical **ENGINEERS** or **PHYSICISTS**

Microwave . . .

to develop test sets and methods on long-distance radio relay equipment. Experienced in performance of amplifiers, modulators, filters and cavities at microwave frequencies.

Filters and Networks to develop manufacturing facilities and methods of manufacture.

Test . .

to develop test facilities for the manufacture of communication and television transmission. Basic knowledge of transmission theory required.

Minimum Qualification B.S. Degree or Equivalent

Write Dept. E giving details of qualifications, experience and salary expected, or

Apply in Person Employment Department Monday through Saturday, 8:30 AM to 3:30 PM

Applicants must furnish proof of citizenship and social se-curity number. Veterans must present discharge papers.

WESTERN ELECTRIC CO.

100 Central Ave., Kearny, N. J.

SUPPLY UNIT of the BELL SYSTEM

A CHALLENGE

to Outstanding **ENGINEERS PHYSICISTS DESIGNERS**

> Take Advantage of the Present to Assure Your Future

POSITIONS with SCOPE and IMAGINATION NOW OPEN

Minimum Requirements Are:

Four years experience in advanced re-search and development on Radar Systems, Computers, Wave Guide and Antennas, Fire Control, Moving Target Indication, Servomechanisms, Pulse Techniques, Gyro-scopic Equipment and Related Fields.

Please send resume and salary requirements to:

The W. L. MAXSON CORPORATION

460 W. 34th St., NEW YORK 1, N. Y.

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Research Laboratories of organization engaged in equipment development, process engineering and fundamental research has positions available as follows:

- 1. Senior Physicist or Physical Chemist for theoretical and experimental work on the behavior of fine particles in fluid media. Must be good experimentalist.
- 2. Senior Physicist having practical experience in the field of electronics and servomechanisms. Must be capable of completely developing electronic devices from the theoretical design through to a finished product.
- 3. Junior Physicist having a strong background in theoretical work as well as possessing experimental aptitude.

Please give education, experience, personal details and salary requirements in first letter.

THE SHARPLES CORPORATION 2300 Westmoreland St. Philadelphia, Pennsylvania

Reply to

THE SHARPLES RESEARCH LABORATORY 424 West 4th St. Bridgeport, Pennsylvania

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Manufacturers Representative Growing organization has capacity for another product in the electro-mechanical field. An established sales and engineering service of unusual quality.

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Burbank, Calif.

POSITIONS WANTED

GRADUATE E. E. with heavy technical and administrative experience in radio and electronic engineering seeks position as chief engineer of small or medium-sized company or assistant chief of larger. Will also consider position as technical group leader. Within commuting distance of Paterson, N. J. PW-9965, Electronics.

EXECUTIVE ENGINEER, extensive background in electronic measurement and control, wants challenging position. Profit-sharing incentive preferred. PW-9973, Electronics.

DEVELOPMENT AND design engineer. 10 years engineering experience in Search and Fire Control Radar, Pulse, Telemetering and microwave circuit design. U. H. F. and microwave antenna development. PW-9817, Electronics.

ELECTROMECHANICAL PROJECT Engineer —30, B.S., M.S., 10 years of experience in development of automatic control systems (servos), precision instruments and production. Aggressive. Desires responsible position with progressive company. PW-9979, Electronics.

GRADUATE E.E.—2 yrs, design of precision electronic meas, mach, in crystal factory, 2 yrs, as senior elec, design engr. for resistance welders and 6 yrs, communications. What I don't know, I catch on quickly. PW-9998, Electronics.

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Ph.D. AVAILABLE for short-term overseas work. Eight years electronic research and development. Unusual knowledge of components and construction techniques. PW-9922, Electronics.

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LINES WANTED, electronic and electrical components and equipment. Missouri and Southern Illinois. Established representative. RA-1023, Electronics.

PATENTS

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Experienced in Service, Development, Electronics & Automatic Machines. Ability to Inspire Employees & Direct Development.

35 Years Old

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Thoroughly experienced in all phases of govt, and comm. sales engineering and management. EE. Seeks new connection on salary or participating basis.

SA-1020 ELECTRONICS

330 W. 42nd St. N. Y. 18

electronic and mechanical engineers

Organization established in 1942. Electronic research, development, and production must expand to meet long-term, pre-Korea commitments. Openings in all branches of electronics, including

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ANTENNAS RADARS
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Positions available for men with at least 5 yrs experience

Write or telephone Howard J. Gresens at

Airborne Instruments Laboratory

160 Old Country Road, Mineola

Garden City

Long Island, N. Y.

Garden City 7-6880

Admiral Electronic Engineers

Needed Immediately for Long Range Research Program

Admiral Corporation, one of America's largest electronic and appliance manufacturers, needs electronic engineers for long range research program covering both consumer as well as governmental projects. Project and senior project engineers with minimum of two to five years' experience needed immediately. Top pay with excellent opportunities for advancement. Send resume of experience and education to Engineering Personnel Manager...

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Dept. E, Chicago 47.

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300	ohi	n twir	lead	. 300)' min	imum	per	ft		.03
Sign	ma 1	olate	relay	8000	ohm-	-SPI)T		٠	2.75



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44	6.4	41/2"	x5"	70.0	18B4J1S4	54.50
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4.4	1.6	23/0"	V:33 0"	3.2	40B4FW1S1	9.75
**	4.4	41/0"	V.5.	6.0	40B4K1S1	14.95
4.4	1.4	5"	×6"	9.0	40B4JW1S1	17.50
6.6	6.6	41/2"	x5"	12.0	40B4K1S2 40B4JW1S2	28.75
4.6	4.4	5"	x6" x5"	18.0 25.0	40B4JW152	42.50
4.5	6.6	4 1/2"	x5"	34.0	40B4J1S4	54.25
14	4.6	41/2" 5"2"	x6"	36.0	40B4JW1S4	62.25
0-120	0-100	13/4"	x13/4"	0.60	40B4D3S1	12.45
0-120	0.100	2-3/16	"×2+3/16"	1.2	40B4EW3S1	14.99
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44	4.4	41/2"	x5" x6"	6.0 9.0	40B4K3S1 40B4JW3S1	32.50
**		5	CENTER		40040 11001	42.50
10-0-	10 0-8	13/4"	×13/4"	1.2	20C2D1S1	2.35
	14	2-3/10	"x2-3/16"	2.4	20C2EW1S1	3.10
1.6	6.6	33/8"	x33/8"	6.4	20C2FW1S1	6.95
6.6	**	4 1/2 "	x5" x6"	16.0	20C2K1S1 20C2J1S1	8.95
4.6	6.4	41/2"	x5"	24.0	20C2K1S2	14.75
44	4.6	41/2"	x5"	36.0	20C2K1S3	19.65
44	4.6	41/2"	x5"	48.0	20C2J1S3	27.50
4.4	4.4		×5"		20C2J1S4	34.50
66	44	4 1/2 " 4 1/2 " 5 "	x5"	80.0		49.50
**	**	41/2"	x5" x6"	84.0	20C2J1S6	55.50
8.4	6.6	41/2"	x5"	128.0	20C2J1S8	67.50
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		13/4"	x134"	0.90	40B6D3S1 40B6EW3S1	24.50
0-120	0.150	0 9/10				
0-120	0-150	2-3/16	x33/e"	4.75	40B6FW3S1	42.50
6.6	6.6	2-3/16 33/8" 41/2"	x33/8″ x5″	9.0	40B6FW3S1 40B6K3S1	42.50 79.50
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Dual Pri	4.25
5v @ 3A; 2.5v @ 2A	5.35
@ 3A; 2.5v @ 2A. 375-0-375 @ 175 MA. 5v @ 3A, 6.3v @ 5A,	4.75
78 v @ 1A. 325-0-325v @ 12 MA: 255-0-255v @ 240 MA.	3.79 4.25
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512.5·0·512.5 @ 427 MA	5.35 2.95 2.95
@ 1.8A 220/440 Pri Step Up/Down 110/220. 500 watt	3.95 10.95

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OIL	CONDENSERS—DC	RATINGS

3X.1	MFD	600v	\$.59	. 1		2500v	
.25	4.4	4.4	.35	.25	1.0	64	1,25
.5	4.6	4.4	.45	.5	4.6	44	1.35
1	6.4	**	.69	2	4.4	4.4	3.45
2	4.4	44	.85	.01	44	3000v	1.25
2 X 2	4.4	4.4	1.15	.05	6.0	4.6	1.30
4	4.4	4.6	1.29	.1	6.4	1.6	1.35
6	4.4	4.4	.98	.25	4.6	6.4	2.75
8	4.4	6.6	1.85	.5	4.4	1.6	2.85
10	**	44	2.25	1	64	4.4	2.95
3X.1	4.6	1000v	.85	2	4.4	4.4	4.25
.5	4.4	6.4	.89	4	4.4	4.6	6.95
1	4.4	4.4	.67	12	6.4	4.6	9.95
2	4.4	6.6	1.75	1	4.4	3600v	3.98
4	1,4	6.6	1.85	.25	4.4	4000v	3.49
8	1.6	6 6	2.45	.5	6.6	6.6	3,75
10	4.4	414	3.19	1	6.6	6.6	3.95
20	4.6	616	4.25	2	4.4	**	5.75
.5	+6	1500v	1.02	3	10.0	4.4	5.89
1	4.4	64	1.19	. 1	9.9	5000v	2.75
2	4.4	4.4	1.69	.25	4.4	4.4	2.95
4	1.4	6.4	2.69	1	64	6.6	3.10
. 1	6 +	2000v	1.75	.1	5.6	7000v	3.75
.25	44	44	1.92	.01	4.4	7500v	2,25
.5	4.4	4.4	1.95	.02	4.6	4.4	2.25
1	4.0	4.4	2.09	.03	4.4	44	2.35
2	4.0	4.4	2.85	.05	4.6	**	2.35
4	4.6	44	4.45	.1	4.6	**	4.95
8	4.4	44	4.95	2X.1	4.6	8.4	7.95
15	4.4	14	6.95	.02	41	12000v	12.95

HIGH CAPACITY CONDENSERS

2.	X8500	MFD	20 V	\$3.47	200	MILD	36V	\$.57
2	500	4.4	3v	.35	100	4.4	50v	.45
3	000	44	25v	2.45	4000	4.4	30v	3.25
6	50	6.6	80v	1.29	2350		24v	2.25
2	000	44	15v	1.69	10000	44	25 v	4.57

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TEST EQUIPMENT CATIONS RADAR— COMMUN

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cyc.
TS-3/AP S-band Frequency and Power Meter. Portable. Battery operated. Complete with all cables.
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put of SCR-718 transmitter. Computer Schless.

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TS-108/AP X-band Dummy Load. Consists of a length of X-band guide filled with sand. One end closed other terminates in a coupling choke. Excellent.

Excellent.

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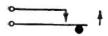
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SHORT TELEPHONE REI	1/-04	3 24 150	1A/50 Amps. 1A/50 Amps.	4.05 R-7	16 24 70 2A/5 Amps.
R-635 12 100 1C & 1A R-648 12 170 1E	1.35 R-71 1.35 R-70	7 24 200 3 12 20	1A/50 Amps. 2A/25 Amps.	3.45 R-6 2.80 R-6	20 6/12 35 2C, 1A 29 9/14 40 1C/10 Amps.
R-826 12 150 2C, 1B R-770 24 150 1A/10 Amps	1.55 R-74	8 24 60	1A/30 Amps. 1A/50 Amps.	2.45 R-7 6.05 R-5	20 24 50 2C, Ceramic
R-538 24 180 2A	1.55 R-6	4 24 150	2A/40 Amps.	2,40 R-8	16 12 10/15 2C/6 Amps.
R-771 24 200 1A/10 Amps R-603 18/24 400 2A	1.55	2 24† 70	1A/25 Amps.	3,05 R-5 R-5	666 115* Coil Only
R-575 24 500 2C R-764 48 1000 2C & 2A	2.40 1.85			R-7	10 150 Coil Only AC † AC/DC
R-563 60/120 7500 1A	1.70 1.45	ROTA	RY RELAYS	_	
R-213 5/8* 2A	3.10 R-7	9 6/8 1	12 Pos. 3 Wafer	\$3.90 1.45	SPECIAL!
R-589 12 125 2A	2.45 R-19 1.30 R-7	3 9/14 125	2C 1C & 1A	2.05	4
R-133 12 150 4A R-689 12/24 255 1C	1.55 R-19 1.55 R-6	6 98 7	3B, 3A, 1C 5B & 1A	2.05 1.90	To The state of th
R-799 24 500 None	1.00 R-7	2 24 200	2B 2C & 1B	2.05 2.05	
R-115 24 500 1C	1.60 R-7 1.70 R-5	1 24 200 3 28 200	1C & 1B	2.05	
R-110 24/32 3500 1C R-121 150 5000 2A & 1C	1.70 R-5 1.70 R-2 2.05 R-2 2.45 R-7	n 94/39 275	4C & IB 3C & IB	1.45 2.05	CUA
R-634 150/250 6000 1A & 1B R-800 12 150 2C & 1A	2.45 R-70 1.55 R-80	66 24 230 19 28 7	12 Pos. 8 Deck. 1B & 12 Pos. W/	4.90	0
R-537 12/24 150 2C & 1B	1.55	.5 26 1	7" Shaft for Wafer	s. 2.45	CO-AXIAL RELAY
R-750 24 400 1A	1.60				D153766 SPDT, 6 VDC. 19 OHI
SLOW ACTING R-547 24 200 1C—Slow Br	reak \$1.80	W-E- TYP	E "E" RELAYS		Designed to accommodate 75 maximum. Perfect for all types
R-548 48 1000 1B—Slow M	ake 2.00 R-7	86 18/24 600	3C	\$1.50 1.30	tenna switching. Designed for
R-128 50 2100 1C—Slow Br R-525 12/24 200 1C/10 Amps	S. S. Mk. 1.55 R-2	37 12/24 650	2C 2C	1.30	standard 83-1SP coaxial fittings of RAX-1 equipment, No. R-845 S
R-794 12 3 Coils. 16 Ohms ea. 3B & 3A/15 Ami	R-1	34 24/32 1000 16 24/48 1300	1A 1B	1.50 1.50	or RAA-1 equipment. No. n-043 5
5 & 5.1/ 10 mm	100	,,			



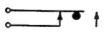


D153766 SPDT, 6 VDC. 19 OHM Coil. Designed to accommodate 75 watts maximum. Perfect for all types of antenna switching. Designed for using standard 83-1SP coaxial fittings. Part of RAX-1 equipment. No. R-845 S6.95 ea.

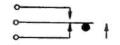
BASIC CONTACT ASSEMBLIES SHOWN IN UNOPERATED NORMAL POSITION



Form A-"Make" (Single Throw, Normally Open)



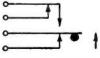
Form B-"Break" (Single-Throw, Normally Closed)



Form C-"Break-Make" (Double-Throw)



Form D-"Make-Before-Break'



Form E-"Break-Make-Before-Break"

WRITE FOR WELLS CATALOG TELEPHONE

SEeley 8-4143



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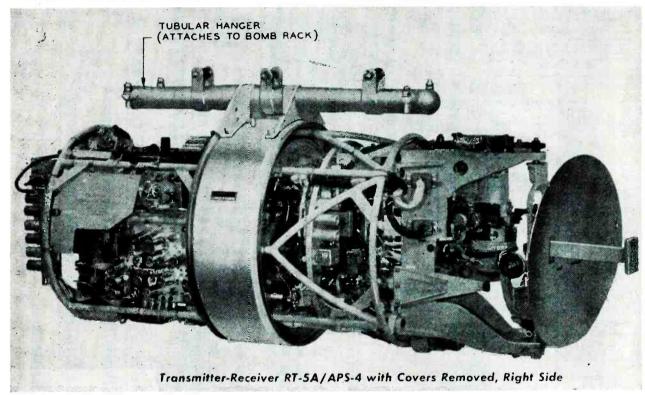
EE8 TELEPHONE PARTS LARGE QUANTITY COILS, BLOCKS, CORDS ARC/5 SPARES & PARTS 2830 KC IFs for 6-9 Rec. (3	CERAMICONS Water Tube 395-475 MMF20 170-315 MMF20 22:195 MMF 20 10 for 15:180 MMF20	#T104 P-P 6L6's to 832	METER SPECIALS AMPS—RF 0-4 Amp RF GE Type DW52. 2% Accuracy. Expanded Scale, 2.19" dia. Int. Thermocouple. List Price \$19.75.	CRYSTALS Low Frequency FT-241. A holder ½" pin spacing, for ham and general use. Xtal con- trolled Signal Generators, marked in army Mc harmonic frequencies—Di- rections for deriving fundamental frequencies enclosed. Listed below by
coils inp. int, out) per set. \$1.25 ### 6234 6-9 RF	10 MFD 600 V 0it Condenser (2.5-2.5-5 MFD) Special Price \$2.95	ILABLE IN STOCK LARGI RC 375 or SCR 191 PAR EQUIPMENT PHONE— SELSYN MOTORS 5VAC 60 oy #C78243 Can used to turn Ant. Small d, Also plaferential C-78249 2"x574". Set of both 2"x574". Set of both "S15.95 PR. MOBILES! C.A.P. C.D.! V. DYNAMOTORS SUPPLY RATINGS OUTPUT INPUT oits 450 VDC 6 VDC Comps 350 MA 45A OR oits 450 VDC 6 VDC Comps 450 VDC Comps 450 VDC 6 VDC Comps 450 VDC Comps .	TS CS	Treduency, fractions omitted: 370 424 472 505 537 372 425 473 506 390 374 426 474 507 391 375 427 475 508 392 376 429 476 509 393 377 430 477 511 394 379 431 479 512 395 381 434 481 514 397 381 434 481 514 397 381 434 481 514 397 383 436 484 516 400 385 437 438 518 386 438 486 519 402 387 440 487 520 388 441 488 522 404 498 388 441 488 522 405 407 412 442 490 523 405 413 443 491 525 406 407 415 415 493 416 458 494 529 409 418 459 495 530 411 418 490 418 459 495 531 490 531 490 531 490 496 531 490 496 531 490 51.49
PL 148 3 png male	D-172155 \$2.25 D-172307 \$1.75 D-167208E \$1.85 D-17268 \$1.85 D-171812 \$1.85 D-171812 \$1.85 D-171828 \$1.85 D-165593 \$1.25 D-162356(308A) \$1.25 D-162356(308A) \$1.50	U.S.A. report ex. re- uit Service of the control of	1.1 600 1ST .22 2.10 1.1 600 2ST .22 2.10 1.3 600 2ST .22 2.10 1.3 600 2ST .22 2.10 2x.25 600 3ST .30 2.85 .05 600 2TT .21 2.00 1 100 2TT .23 2.20 20 50 2ST .26 2.50 VARIABLE TRIMMER CONDENSERS C890 15 MMF C993 50 MMF C993 50 MMF C677 60 MMF C677 60 MMF C677 60 MMF C677 60 MMF	423 470 504 533 534 59c Page 201
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Airborne Search-Intercept equipment—A light weight airborne equipment for operation at frequencies in the X Band designed primarily for use on the Navy's carrier-based planes. It can be employed to perform the following functions on any type

- 1. Radar Search operation over large bodies of water for the detection of vessels on the surface.
- 2. Mapping and navigational operation.
- 3. Detection and location of other aircraft to permit their interception.
- 4. Picking up the coded signals from the beacon stations in the X-band to facilitate the return of the searching plane.
- 5. Operation as the search element in conjunction with identification equipment.



Complete Installation—Packed—Write for Price and Details—"Underbelly Assembly" As Illustrated Less Tubes \$375

This equipment searches by use of a "nodding" and sweeping paraboloid antenna which emits a beam of RF pulses into space and receives reflected pulses back from targets in the path of the emitted beam. The received pulses considerable manages in the path of the emitted beam. The received pulses to gettler with "antenna position" infermation and time they are received, are displayed on indicators in a manner which determines azimuth and "slant range" of targets within the area searched. The antenna can be oscillated to scan 70 degrees of azimuth on either side of dead ahead and can be tilted with the plane to 30 degrees below xias assemble position affords sufficient radiation to some state of the search of the plane and for a considerable distance surrounding that point, which makes it deal for detection of SUBMARINES. Nominal ranges of operation for this equipment are as follows:

(1) Ships 30 Miles (2) Aircraft (3) Land Targets 30 of depressions affords sufficient radiation to search of the sequipment areas follows:

(2) Aircraft (3) Land Targets 30 Miles (4) Aircraft (5) Miles 30 Miles (4) Aircraft (5) Miles (5) Miles (6) Miles (6) Miles (7) Miles

ency of operation is inthe order or 9, 310-9, 375 MCS.

sectiver—Transmitter Unit
RT-5A/3PS-4
The receiver-transmitter is contained within a "bomb-shaped" fibrecomposition housing and is commonly called the "bomb." The bomb-unit
is sealed and pressurized on the ground with dry air to a gage pressure
of 5 lbs. per square inch. It is usually mounted beneath the plane's
lower wing and supported in a standard Kay bomb rack whereby (in
emergency) the pilot can dispose of the unit by activating the release
mechanism. Located in the nose of the "bomb" is the paraboloid reflector and wave-guide fed antenna. The transmitter portion comprises
four major circuits, namely:
(1) Triggering Circuit
(2) Charging Network
The charging network is charged to shigh potential, then rapidly discharged thorugh a pulse transformer which impresses a large negative
voltage of short duration to the cathode of a "grounded-plate" magnetion oscillator which causes the magnetion to vigorously oscillate,
for the duration of its cathode impressed pulse, and send forth to the
antenna a powerful burst of RF energy at a frequency of 9, 375 MC. The
triggering circuit initiates the pulse which causes the discharge of the
network. When the magnetron oscillator "fires" its burst, that energy
passes through a T-R Box enrouse to the antenna. The T-R Box acts
as a valve which functions to isolate the receiver input from the transmitter output, yet when the weaker reflected signals are picked up by the
antenna, this same T-R Box allows that weak signal to be accepted by
the receiver portion undamped by the transmitter portion of the caulpment because of its isolation. The incoming signal, after clearing the
T-R Box, is then mixed with the output from a local oscillator, in a
crystal converter from the output of which is picked off a 60 MC IT
signal. There are two local oscillators, one of which is used during
SEARCH and INTERCEIT operations and is tuned to 60 Mecloring
SEARCH and INTERCEIT operations and is tuned to 60 Mecloring
the T-R Box is

Following this a 6AK5 detector (connected as a cathode follower feeds the video amplifier. Part of the output from the 2-stage preamplifier is used to energize an AFC circuit which controls the frequency of the local oscillator. At the video amplifier (which is a separate unit from the Receiver-Transmitter) the video echo signals are mixed with RANGE MARK PII'S and the combination is applied to cause signal indications to appear on indicator screen which describe the range of the object from which the initial transmitter output signal was reflected.

Tube Complement:

TRANSMITTER PORTION TRANSMITTER PORTION

1 ea. — JAN 1B22 4 ea. — JAN 3B24 3 ea. — JAN 6J6 1 ea. — 725A 1 ea. — JAN 807

 s ea. — JAN 6465
 IF AMPLIFIER PORTION

 6 ea. — JAN 6A65
 RANGE AND ALARM UNIT

 15 ea. — JAN 646
 1 ea. — JAN 6A65

RECEIVER PORTION
2 ea. — 723A/B
2 ea. — 724A 2 ea. — JAN 6J6 3 ea. — JAN 6AK5 1 ea. — JAN 6SL7GT

CONTROL AMPLIFIER PORTION
1 ea. — JAN 6AK5 2 ea. — JAN 6J6

CONTROL AMPLIFIER PORTION

1 ea. — JAN 6AK5

BEACON TUNER PORTION

1 ea. — JAN 6J6

BEACON TUNER PORTION

1 ea. — JAN 6J6

2 ea. — JAN 6J6

3 ea. — JAN 6J6

4 ea. — JAN 6V6GT

6 ea. — JAN 6V6GT

6 ea. — JAN 6V6GT

6 ea. — JAN 6V6GT

7 ea. — JAN 6V6GT

8 ea. — JAN 6V6GT

1 ea. — JAN 3FP7

2 ea. — JAN 6J6

3 ea. — JAN 3FP7

2 ea. — JAN 6J6

4 ea. — JAN 6J6

6 ea. — JAN 6V6GT

7 ea. — JAN 6V6GT

8 ea. — JAN 6V6GT

8 ea. — JAN 6V6GT

9 ea. — JAN 6V

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No. P 1500 2TT D164209 600 2ST RO3. 600 2ST RO3. 400 2BT 616M 600 2BT 616M 600 2BT 7700BR 300 3BT NJOCA195. 340 3BT NJOCA195. 340 3BT NJOCA195. 340 3BT R1-616MB 400 2BT 616M-14842. 600 2BT 616M-14842. 600 2BT 770BR 600 2BT 780BC 600 3BT 770BR 600 3BT 70BBC 600 3BT CD516 600 3BT CD516 AM 1510M 7 Amp 1614 80 Amp 1614 50 Amp Mfd. Volt Type Cat. No. Price 1.45 AT353 2.95 AT871 Price each\$1.10 THE CONDENSERS LISTED AT554 AT765 BELOW ARE NEW, UN-MICA COND. AT707 AT449 MARKED, TESTED FOR CA-XM Screw 2000 55d 3.89 PACITY. MFG BY SOLAR AT21 2550 55¢ Type G 3000 55¢ 5000 VT AT383 AT415 AT649 3450 XO Lug XO Lua 55₫ 15 4000 60¢ Type E Type E 21 456 4800 60¢ 1200 VT .1 2x.1 3x.1 3x.1 3x.1 3x.1 3x.1 2x.125 .25 5000 VT 50 450 5000 60¢ OIL CONDENSERS Price MMF 140 456 5900 600 Mfd. MMF. MME MMF 235 456 6900 650 5 650 3.1 4 11 13 14 15 20 220 25 250 45é 7000 656 28 35 58 85 900 3BT 710B1 42¢ 400 2TT 400 2TT 690 800 2WTT 690 800 2WTT 600 2BT C168B1EF254K 30¢ 600 2BT C168B1EF254K 30¢ 600 2BT S01 32¢ 600 2BT S01 32¢ 600 2BT S01 32¢ 400 2TT 416T 30¢ 600 2BT S01 32¢ 400 2TT 416T 30¢ 600 2BT 818MCB 32¢ 400 3BT 18MCB 33¢ 400 2TT 085 35¢ 600 2BT 71 61MCB 33¢ 600 2BT 71 61MCB 33¢ 600 2BT 71 61MCB 33¢ 600 2BT 61MCB 35¢ 150 AC 220 AC 500 295 456 5 15 50 51 60 62 65 70 8500 65d ₹ 300 456 10000 70¢ .40 1.25 2.10 .69 .70 .75 .90 1.75 1.35 1.40 1.45 2.25 2.45 1.35 1.20 1.20 .25 .4 .25 .5 .5 .5 .5 .2x.5 2x.5 2x.875 375 456 10500 70¢ 20 0.5 0.5 2x0.5 AC 400 45é 12500 70¢ 1000 490 45¢ 40000 1000 1000 1000 1000 1200 1500 \$7 PER 100 500 456 1.5 SILVER MICA BUTTONS 575 456 XM Screw MMF. | MMF | MMF. 40 | 175 | 500 50 | 180 | 2000 185 | 57.50 PER 100 750 456 3x.01 0.0016 Type G 850 45¢ 7500 VT 1000 50¢ 1.5 1500 1500 1150 50¢ HEADSETS Dynamic Mike and Headset Combination. A high quality, efficient unit, used in B-19 tank Xmtrs. Mike and phones complete new. \$3.75 50d 750 2x0.1 2000 2000 2000 2000 2200 1250 50¢ 850 50¢ 1.75 1500 1000 5**5**¢ 0.5 1.5 0.15 2x0.1 0.1 2x0.1 0.1 WRITE FOR CATALOG VG 1700 1500 55¢ 50¢ 604 XM Screw OF 1,000,000 CONDENSERS 1850 2000 4000 4800 6000 7000 1500 1850 2000 2250 2300 Type D 1.20 1.20 1.25 1.25 4.75 400 CYCLE TRANSFORMERS TOP TRANSFORMER BUYS 1200 VT These XFRMRS are Army Spec. 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Transformers—115V/50-60 cps input iV/50-60 cps Price 7.5VCT/4A, 2.5V/2A, 6.3V/1.2A, 4.2S 6.3VCT/3A, 3.9S 2 x 12.6V/10. 1.9S 2.5/12. 30/100. 9.9S 6.3/6.6.6.3VLT/3A, 3.9S 2.5/12. 30/100. 9.9S 6.3/6.6.3/1.8 3 lbs 2.9S 6.3/6.6.5/1.8 3 lbs 2.9S 6.3/6.6 6.3/1.8 3 lbs 2.9S 6.3/6.8 6.3/6.8 3 lbs 2.9S 6.3/6.8 6.3/6.8 6.3/6.8 3 lbs 2.9S 6.3/6.8 6.3/6. Each \$1,49 3.95 3.49 2.49 1.95 2.49 2.99 3.95 5.95 MANY OTHERS Item H.V. Amp. CT-861 2100VCT 175 Tuning Units for BC-375 & 191 Diag. for Conver. to VHF Freq. Meter in Oct., CT-142 645VCT CT-825 360VCT 850 1300 1700 2900 4000 4200 4800 5400 5500 6700 6900 8500 11500 11500 32000 335000 35000 35000 CT-076 600V CT-626 1500V CT-15A 350VCT CT-071 110V 100 .160 .070 .200 CT-076 600V 100 2 x 12.6V/1 1.95 CT-1526 1500V 160 2.5/12. 30/100. 9.95 CT-15A 350VCT .070 6.3/.6, 6.3/1.8 3 lbs 2.95 CT-071 110V .200 33.200, 5V/10. 4.95 CT-367 580VCT .050 5VCT/3A .2.25 CT-721 550VCT .100 6.3/1, 2.5VCT/2A .2.5 CT-721 550VCT .100 6.3/1, 2.5VCT/2A .2.5 CT-93A 2x110VCT .010 6.3/1, 2.5VCT/2A .2.5 CT-94A 2x110VCT .010 6.3/1A, 2.5VCT/2A .2.5 CT-94A 550VCT .026MA 5V/3A, 6.3/3.5. 3.25 CT-44B 50V .200 5V/2.4, 5V/12. 2.29 CT-401 550VCT .026MA 5V/3A .2.75 CT-931 585VCT .026MA 5V/3A .2.75 CT-610 1250 .002MA 2.5V/2.1A, 2.5VCT/16 .4.95 CT-102 1080VCT .055 6.3V/1.2A .4.95 CT-1319 330VCT .085 5V/3A, 6.3V/1.8A ,5.95 G3V/1.2A ,5.95 Filament Transformers—115V/50-60 cps input THT 4500-62000KC TU8 6200-- 7700KC TU9 7700-10000KC 2.95 4.95 2.95 .98 1.49 TU26 200-500KC Price \$2.49 ea. 6,49 3,49 3,25 2,25 1,49 Type E 2500 VT 300 3.95 2.95 35 50 **DYNAMOTORS** 30d 100 130 Input Volts Ar 300 XM Screw Туре Item Rating FT-029 3.5V/1.11A. 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26

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MOTOR GENERATORS DYNAMOTORS, INVERTERS, ETC.

2.5 KVA MG Set. Diehl Elec. Co. 120V D.C. to 120V A.C. 60 cy. 1Ph., .4PF. Complete with Magnetic Controller, 2 Field Rheos and Full Set of Spare Parts including Spare Armatures for Generator and Motor. Full specs. on request. New \$285.00

2 KVA MG SET. O'Keefe and Merritt, 115V DC to 120V AC, 50 cy. Idles at 3 Ph. syncs motor on 208V, 50 cy. New. Export crated \$165.00

1.25 KVA MG SET. Allis-Chalmers. 230 DC To 120 AC, 60 cy. 1 Ph. Fully enclosed, Splashproof Ball Bearings, centrifugal starter, New, complete with kit of Spare Parts...\$150.00

IG Set. Onan MG-215H. Navy type PU/13. Input 15/230, 60 cy, 1 Ph. Output 115, 480 cy, 1 Ph. 200W and 26V DC at 4 amps. New.....\$295.00

DYNAMOTOR. Navy-Type CAJO-211444. 105/130V-DC to 13V DC at 40A or 26V DC at 20A. Radio filtered. Complete with Line Switch. New. \$69.50

DYNAMOTOR. Elcor. 84V DC to 110V AC, 80 cy. 1 Ph, 2.04 Amps. New......\$24.50

INVERTER—G.E. Model 5D-21NJ3A. Input: 24V. DC, Output; 115V, 400 cy. 485 Va. New...\$24.50

PE 218 INVERTER—G. E. J8169172. Input: 28V. DC, Output: 115, 400 cycles at 1.5 KVA...\$24.50

GENEMOTOR—Carter 8V DC to 400 V DC at 375 mils. New\$39.50

D.C. MOTOR—G.E. Model 5BA50LJ2A. Armature 27V D.C. at 8.3A. Fleid 60 V D.C. at 2.3A. RPM 4000. H.P. 0.5. New \$22.50

MOTOR GENERATOR M.G. 164. Holtzer-Cabot Motor: 440V, 3PH, 60 cy., 90A, 1/3HP, 1750 RPM. Generator: 70V, 3Ph, 146 Cy., 140KVA Exciter: 115DC, 1A. New. \$67.50

BC-348 RECEIVER PARTS

Dial Mechanism assemblies. 1st, 2nd, 3rd, 4th I.F. transformers. C.W. osc. and xtal filter trans. with xtals. All R.F. coils. Front panels. Shock mounts. Large quantity misc. hardware sub assemblies, etc. Write your requirements.

MISCELLANEOUS EQUIPMENT

TS-127/U Lavoie Freq. Meter—375 to 725 MC. TS-47APR Test Set—40 to 500 MC. 213-A DuMont C.R. Modulation Monitor. BC1203B APN-4 Test Set, 6255A H.P. Interpolation Osc. TS-23/APN Test Set. TS-487/U Peak to Peak VTVM. BC-221AE Freq. Meter.

MICROWAVE RECEIVERS

AN/APR-1 Receivers and tuning units TN-1 (38 to 95 MC) TN-2 (76-300 MC) TN-3 (300-1000

MODEL AN/APA-10 PANORAMIC ADAPTER



Provides 4 Types of Presentation: (1) Panoramic (2) Aural (3) Oscillographic (4) Oscilloscopic

Designed for use with receiving equipment AN/ARR-7, AN/ARR-5, AN/APR-4, SCR-587 or any receiver with I. F. of 455kc. 5.2mc, or 30mc.

SUMMARY OF CHARACTERISTICS:

SENSITIVITY: "A" channel, 400 microvolts or less per %" beam deflection, "B" channel, 400 microvolts or less per %" beam deflection. "C" channel, 1 volt or less per %" beam deflection.

RESOLUTION: 12 kilocycles at 3 db down from peak, sweep control at maximum, using CW signal.

PRESENTATION: Panoramio ("A" & "B" channels): Oscillotraphic, "C" channel.

SWEEP WIDTH: Channel A. ± 50 kc (100 kc overall) Channel B, ± 500 kc (1 Mc overall) Channel C, ± 1 Mc (2 Mc overall).

CATHODE RAY SWEEP: Oscillatory or non-oscillatory (Servo) Variable Sawtooth Generator, 35 to 40,000 cycles per second.

AUDIO OUTPUT: 50 milliwatts into 600 or 8000 ohm load.

VERTICAL AMPLIFIER: Single stage ± 2db from

ohm load.

VERTICAL AMPLIFIER: Single stage, ± 2db from 30 cycles to 100 ke or higher. Amplifier out position permits direct connection to one vertical plate through coupling capacitor.

HORIZONTAL AMPLIFIER: Single stage, ± 2db from 30 cycles to 100 ke. No provision for direct connection to deflection plates.

CATHODE RAY TUBE VOLTAGE: Cathode to accelerating anode; 1200V DC for 115V A.C. input.

SENSITIVITY OF CATHODE RAY OSCILLOSCOPE: Maximum through Amplifier. Horizontal: 10 volts peak to peak per inch. Vertical: 1.5 volts peak to peak per inch.

peak to peak per inch.

DIRECT TO VERTICAL PLATE: 150 volts peak to peak per inch.

NOISE: No disturbance in excess of 25,000 microvolts between 200kc to 200Mc generated by equipment. Overall Dimensions: 19-9/16" x 1034" x 75%".

Weight: 40 lbs.

Power Requirements: 115V. A.C. 60 cycles, 1 phase. With 21 tubes including 3" scope tube, for operation on 115 V. 60 cycle source. PRICE......\$245.00

AN/APA-10 80 Page Tech Manual.....\$2.75

LINEAR SAWTOOTH POTENTIOMETER W.E KS-15138

Has continuous resistance winding to which 24 volts D.C. is feel to two fixed taps 180° apart. Two rotating brushes 180° apart take off linear sawtooth wave voltage at output.

Brand New \$5.50

G. E. SERVO AMPLIFIER

Type 2CV1C1 Aircraft Amplidyne control amplifier, 115 volts—400 cycles. Dual channel. Employs 2-6SN7GT and 4-6V6GT tubes. Supplied less tubes. New \$22.50

400 CYCLE TRANSFORMERS

AUTO. 400 cy. G.E. Cat No 80G184. KVA .945S—520P. Volts 460/345/230/115. New. \$4.95

FILAMENT. 400/2600 cy. Input: 0/75/80/85/105/115/125V. Output: 5V3A/5V3A/5V3A/5V8A/5V6A/5V6A/5V6A/5V6A/65V6A

1500V)/5V2A/5V2A RETARD. 400 cy. WECO KS9598, 4 Henry 100MA \$1.75

60 CYCLE TRANSFORMERS

PULSE TRANSFORMERS

PULSE. WECO KS-9563. Supplies voltage peaks of 3500 from 807 tube. Tested at 2000 Pulses/sec and 5000V peak, Wdg. 1-2=18 ohms. Wdg. 1-3=70 ohms. L of Wdg. 1-3=.073-.082H at 100 cps. \$5.00 PULSE. WECO KS-101310, 50 KC to 4MC. 1%" Dia. x 1%" high. 120 to 2350 ohms, New . \$1.95

RAYTHEON VOLTAGE REGULATOR

Adj. input taps 95-130V., 60 cy. 1 Ph. Output: 115V. 60 Watts, $\frac{1}{2}$ of 1% Reg. Wt. 20 lbs. $6\frac{1}{8}$ H x $8\frac{1}{8}$ W. Overload protected. Sturdly constructed. Tropicalized. Special.\$14.75

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SOUND POWERED PHONES

Western Electric No. D173312, Type O. Combination headset and chest microphone. Brand new Including 20 ft. of rubber covered cable...\$17.50 Automatic Elec. Co. No. G1843AO. Similar to above but including Throat microphone in addition to chest microphone. Brand new with 20 ft. rubber covered cable...\$10.00 covered cable \$10.00
U. S. Instrument Co. No. A-260. Complete with
20' cable and plus. Brand new \$13.50
W. E. type TS-10M Handset. New \$16.50

PARABOLOIDS

Spun Magnesium dishes 17½ dia 4 deep. Mounting brackets for elevation and azimuth control on rear. 1½ x 15% opening in center for dipole, Brand new, per pair. \$8.75

SWEEP GENERATOR CAPACITOR

High speed ball bearings. Split stator silver plated coaxial type 5/10 mmfd. Brand new.......\$2.50

WESTERN ELECTRIC CRYSTAL UNITS

Type CR-1A/AR. Available in quantity—following frequencies—fundamentals. 5910—6850—6870—6470—6510—6610—6670—6690—69940—7270—7350—7380—7390—7480—7580—9720—Killocycles.

\$1.25 each

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TIMING MOTOR 8 RPM 1115V 60 cyc E. Ingraham Co



\$1.95

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VERNIER DIAL or DRUM (From BC-221) DIAL—25% dia. 0-100 in 360°. Black with silver marks. Has thumblock. DRUM—0-50 in 180°. Black with silver marks. —either, 85c

SOUND POWER HANDSET



Brand New!

Includes 6 ft. cord. No batteries or external power source used. \$17.60 pr.





Variac—General Radio 100W removed from equipment

\$10.00

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400 CYCLE INVERTERS Leeland Electric Co.

AMP	Per 100 \$4.00 4.00 \$4.00	1 3.00 1½ 3.00	6 \$3.00 10 3.00
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DELAY NETWORK-ALL 1400&

т	113-Approx. 1.2	micro sec. delay	95 <i>č</i>
T	114—Approx. 2.2	micro sec. delay	each
т	115 Similar to T	114 with tap brought out	

BEARINGS

Mfg. No.	ID		Thickness	Price
MRC5028-1	5 1/2	6 1/2"	1"	\$3.50
MRC7026-1	5 5/64	6 15/64	9/16	3.50
Timken 37625	4 5/16	6 1/4	29/32	4.25
MRC-7021-200	4 1/8	59/32	23/64	2.95
Fafnir B545	2 1/16	2 5/8	1/4"	1.00
MRC 106 M2	1 17/64	27/16	25/64	1.75
MRC 106 M1	1 13/64	2 7/16	25/64	1.60
Federal LS 11	1 1/8	2 1/2	5/8	1.75
Norma S 11 R	1 1/8	2 1/8	3/8	1.25
Fafnir B 541	1 1/16	1 1/2	9/32	.55
Hoover 7203	5/8	1 9/16	7/16	.90
Norma 203 S	5/8	1 9/16	7/16	.90
Schatz	3/4	1 3/4	9/16	1.00
N5 5202-C13M	1/2	1 3/8	1 3/8	1.00
ND 3200	25/64	1.5/32	11/32	.55
Fafnir S 3K	3/8	7/8	7/32	.45
MRC 39 R1	11/32	1 1/32	5/16	.45
ND CW 8008	5/16	5/16	13/32	.45
MRC 38 R3	5/16	55/64		.45
Fafnir 33K5	3/16	1/2	5/32	.25

	NEE	DLE	BEA	RINGS	
TORRINGTON	B108	1/2"	wide	5/8"	13/16"

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	D	Manue	METERS-	Guara	boot
0 1	Amn Di	F 914"	\$3.29 L 0-80	Amp DC	21/6" \$2.2

0-1 Amp. R.F. 21/2" \$3.29 0-80 Amp. D.C. 21/2" \$2.25 0-7.5V. A. C. 31/2"... 3.46

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SPAGHETTI SLEEVING-assortment-99 feet... TYPE "I" POTENTIOMETERS

150 300 300 400 500 1000 1000 1500	SD 3/8 3/8 3/8 3/8 7/8 SD* SD	2000 2000 2000 2500 2500 3000 4000 5000	SD 1/2" SD* 1/2" SD 3/8 3/8 SD*	5000 10K 10K 15K 15K 25K 25K 30K	3/8" 3/8" SD* 1/4" SD* 3/8" SD* SD*	50K 70K 80K 100K 200K 250K 500K 1 Me	-
* Split locking bushing \$1.25 e							

JONES BARRIER STRIPS

Type 2-140 Y 3-140 ¾ W 6-140 10-140 ¾ W 2-141 3-141 ¾ W	Price \$0.13 .19 .25 .53 .13	Type 4-141W 5-141 5-1411/W 7-141 7-1413/4 W 8-1413/4 W	Price \$0.30 .26 .37 .36 .49	Type 9-141Y 10-141 17-141Y 3-142 8-142 2-150	Price \$0.64 .50 1.17 .21 .69
3-141W	. 24	9-141	.64	3-150	.54



TIME DELAY RELAY

Raytheon CPX 24166 KS 10193-60 S
115 V. 60 cycle • Adj. 50-70 Seconds
2½ second recycling time—spring return
Micro-switch contact, 10A • Holds ON
long as power is applied • Fully cased ONLY

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IMMEDIATE SERVICE PHONE! WIRE! WRITE! YOUR NEEDS

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RG-15/U 76	160	RG-39/U	72.5	180
RG-21/U 53	100	RG-41/U	67.5	295
RG-22A/U*95	150	RG-54/U	58	65
RG-24/U 125	240	RG-55/U	53.5	65
RG-25/U 48	575	RG-57/U*	95	100
RG-26/U 48	75	RG-58/U*	53.5	60
RG-27/U 48	290	RG-59/U*	73	70
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	irders		,000 feet	
*No minimum o	order	others 250'	minimum	



DIFFERENTIAL

115 V., 60 Cyc. \$3.95 ea,



be Co M di

2J1G1 SELSYNS \$2.95 BRAND NEW 400 CYCLE

Can be used on 60 cycle

POSTAGE STAMP MICAS

MMF	MMF	MMF	MMF	MMF	MFD	MFD
8.2	40	90	240	510	.0011	.0051
10	43	100	250	560	.0012	.006
15	47	110	270	580	.0013	.0062
20	51	120	300	600	.00136	.0065
22	56	125	350	620	.0015	.0068
24	60	130	370	680	.001625	.007
25	62	150	390	800	.002	.0075
26	68	160	400	820	. 0026	.008
30	75	175	430	910	.0027	.0082
35	82	180	470	MFD	.003	.009
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8.5	2 MMF	to .001	MFD											. 5
.001	1 MFD	to .001	$525 \mathrm{M}\mathrm{FI}$)					٠	٠				3
.002	MFD	to .009	MFD.						٠					114
.01	MFD						٠			٠	٠			. 2.

SILVER MICAS

MMF	MMF	MMF	MMF	MMF	MMF	MFD		
10	51	120	270	470	875	.00282		
18	60	125	300	488	900	.00282		
20	62	150	325	500	MFD	.003		
22	66	170	360	510	.001	.0033		
23	68	180	370	525	.0011	.0039		
24	75	200	390	560	.001625	. 00 5		
30	82	208	400	570	.0022	.0051		
39	100	225	410	680	.0023	.0056		
40	110	240	430	700	.0024	.006		
50	115	250	466	800	.0026	.0082		
Price Schedule								
10 MMF to .001 MFD								

FILAMENT TRANSFORMER

Pri., 115V., 60 Cyc. — Sec. $\begin{cases} 6 & \text{V. } @ 35 \text{ A.} \\ \text{or } 12 & \text{V. } @ 18 \text{ A.} \\ \text{or } 24 & \text{V. } @ 9 \text{ A.} \end{cases} \textbf{$6.50}$

PULSE TRANSFORMERS

UTAH—9262 9278 9340
WESTERN ELECTRIC—D166173 D161310
KS8096. KS9365. KS9800. K89862. KS13161
GENERAL ELECTRIC—K2731. 80-G-5
JEFFEIXON ELECTRIC—C12A-1318
CROSLEY—W-226262-4
DINION COIL—TR1048. TR1049
also 352-7250-2A: 352-7251-2A: T-1-229621-60

COAXIAL CABLE CONNECTORS



45.	0.4

3.	e	
į	45¢ 83-1R	9¢ Н ОО D
	UG58/U UG59A/ UG60/U UG85/U UG87/U UG88/U UG167/	U 2.25 2.40 .88 .79 1.25
,	00107/	2.00

0¢ -1J	45¢ 83-1R	9¢ н оо р
10	UG58/U UG59A/	\$.63 U 2.25
63	UG60/U	2.40
95	UG85/U UG87/U	.79
67	UG88/U UG167/V	
67	UG175/1 UG176/	J .15
00	TICONO /	

1	e Com	1
ė 1 J	45¢ 83-1R	9¢ Н ОО Д
3 5 7 0	UG58/U UG59A/ UG60/U UG85/U UG87/U UG88/U	U 2.25 J 2.40 J .88 J .79 J 1,25
0	UG167/	U 2.00

•7 I	UG88/U	1.25
0	UG167/U	2.00
7	UG175/U	.15
50	UG176/U	.15
58	UG201/U	2.05
15	UG206/U	.63
3	UG281/U	.60
50	UG290/U	1.15
30	UG499/U	1.25
	-	

13/8" dia. x 53/8" long	
sed between two #C78248's as dampener. Car	ı.
e converted to 3600 RPM Motor in 10 minutes	
onversion sheet supplied. (Converted) \$4.50	9
ounting Brackets - (Bakelite for selsyns, and	1
ifferentials shown above	Γ





1	1



2 WATT

PRECISION RESISTORS

CISION RESIST
1/4 WATT—30c
12.32 16.37
13.02 32
13.52 62.54
13.89 79.81
14.98 105.8
15.8 123.8
1/2 WATT—30c
120 855

855 900 970 1,060 1,100

200 204 375 400 507 573 876 236 236 236 236 236 237 850 3,387 400 4,000 4,000 4,000 4,000 4,000 4,000 4,000 4,000 4,000 5,714

5,714 TT—35c 312 321.7 400 420 425

565 800 1,000 1,500 1,500 1,800 2,000 2,215 2,250 2,413 2,500 3,055

13. 3 15. 52 18. 75 20. 5 21. 5 30. 75 40. 44. 73 44. 73 44. 73 44. 73 44. 73 46. 66. 6 66. 6 67. 71 75. 87 90. 87 97. 8

.66

4.7 5.21 10.1 10.5 12.8 15 18 25 27.4 28 30 35

5,000

 $01 \\ 02$

.7 .87 .24 .26

6

10.58 11.1 13.15

 $\begin{array}{c} .1 \\ .11 \\ .147 \\ .2 \\ .25 \\ .31 \\ .4 \\ .861 \end{array}$

.01 .166 .17 .21

23,000



		OIL F	ILLED		
MFD.	V.D.C.	Price	MFD.	V.D.C	Price
.125	35,000	\$34.95	.25	2,000	1.05
.125	27,000	28.95	.11	2.000	.95
.25	20,000	18.95	.02	2.000	.70
.05	16,000	2.95	2	1,500	1.75
.03	16,000	1.95		1.500	1.50
.1	12.000	1.95	12	1.000	2.45
1	10,000	9.75	6	1,000	1.79
.1515	8,000	2.45	4	1,000	1.39
.02	8,000	1.69	3	1.000	.80
.05	7,500	1.50	1.02	1.000	.65
.03	7,500	1.40	1.02	1,000	.59
.11	7,000	1.69	1	1.000	.59
.0203	7,000	1.35	7	800	1.59
1	6,000	5.25	1	800	1.49
.0303	6,000	1.35	8-8	600	1.60
.03	6,000	1.25	8 7	600	1.40
.0103	6,000	1.35	7	600 600	1.30
2	5,000	4.65	-5	600	1.25
.25	5,000	2.25	4	600	1.95
3	4,000	4.50	4x3	600	.45
3 2 3x.2	4,000	3.95	2	600	.39
3x.2	4,000	2.35	.5	600	.35
.1	4.000	1.55		500	1.45
.06	4.000	1.35	8	500	.89
.25	3,000	1.50	8 4 4	300	.59
.1	3,000	1.45	4	200	39
4	2,500	4.25	4	200	147
8	2.000	4.45		2 m	fd
6	2,000	4.25	CAA	4.0	
4	2,000	3.65			
3 2	2,000	3.50	(E)	V.D	.C.
2	2,000	2.95		G.I	E.
	2.000	1.95	4		
.5	2.000	1.60	13	\$3.	95
.3	2.000	1.35			

1N31 Crystal Diode

Minimum Crders \$3.

Attorders f.o.b. PHILA. PA

E MERCHANDIZING CO.

Arch St. Cor. Croskey Phila. 3, Pa. Telephone Rittenhouse 6-4927

METERS

3 MA DC 21/2" R-Simpson black scale\$	3.35
500 Microamps, DC-21/2" round-Sun	4.30
ma. DC Fan type-4" scale (rem. from equipt)	
500 ma. DC 21/2" R.—General Electric	2.95
2 amp. RF 21/2 Sq.—Simpson 3	
5 amp. AC 4/2 R.—JBT	
10 amp. RF 31/2" R.—Simpson	4.95
50 amp. AC 31/2" R.—General Electric	4.[[
3 amp. RF 31/2 R.—Weston	5.00

MAGNETRONS

			-
2J21A	2J36	2J61	706CY
2J22	2J37	5J23	706FY
2J26	2338	5J29	706GY
2J27	2J39	700B	714AY
2J31	2J40	700C	718AY
2J32	2J41	700D	720B/C/DY
2J33	2J48	706AY	725A
2J34	2J49	706BY	730A

KLYSTRONS

2K23	2K33	707A	726A
2K25	2K54	707B	726B
2K26	2K55	723A	5611
2K29	417A	723A/B	

OIL-FILLED HIGH VOLTAGE ISOLATION TRANSFORMERS

Pri. 460V 60 cy, Sec. 115V 200VA Insulated for 50KV DC—G, E. Form E1R—36"H x 13"D....\$125.00 Pri. 115V 60 cy, Sec. 115V 250VA Insulated for 35KV DC—G, E. Form E1R—29"H x 12½"D...\$125.00

VOLTAGE DIVIDER

20 LOW INERTIA SERVO MOTORS

KOLLSMAN Type 936-0240—85/68V 100 cy 5 watts 2650 RPM—new \$12.95 DIEHL Type FPE-25-11 75V 60cy 4 watts new \$34.50

OIL FILLED CONDENSERS

MFD	VDC		Price	MFD	VDC	Price
2	600		\$.45	.1	2500	.49
4	600		1.25	1-1	2500	3.85
4	600	R'd	1.25	32	2500	15.80
6	600		1.65	3x.2	4000	2.95
8	600	R'd	1.39	1	5000	4.88
10	600	R'd	1.52	.0103	3 6000	1.65
8-8	600		1.49	.1	7000	1.79
i	1000		.62	2	12500	28.95
2	1000		.89	.045	16KV	4.70
4	1000		1.65	.05	16KV	4.95
8	1000		2.45	.075	16KV	8.95
1	1500		.89	.25	20KV	18.95
4	1500		2.95	50	220VAC	3.95
.15	2000		.87	7	660VAC	4.25
1	2000		1.50	8	660VAC	4.50

HIGH VOLTAGE TRANSFORMERS

Sec. 6250V 80 MA—12.5	KV Ins\$18.50
G. E Pri 115V 60 cy. Sec.	6250//3850/2600V 56 MA
12.5 KV Ins	Sec. 9500/6450V CT 43
MA Hermetically sealed.	\$22.50

CRYSTAL DIODES

1N21B 3.75 1N23B 5.25 1N45 .94	1N21A 1N21B	1.69 3.75	1N23A 1N23B	3.25 5.25	1N38 1N45	\$.79 1.66 .94 1.05

ANTENNAS

AT-38A/APT (70 to 400MC)	\$13.70
AT-49/APR-4 (300 to 3300MC)	
DZ-2 Loop antenna with pedestal	14.50
AN-74B (125 to 150MC)	1.65
AN-65A (P/O SCR-521)	.95
AN-66A (P/O SCR-521)	1.15
AIA-3CM conical scan	125.00
ASB Yagi-5 element 450 to 560MC	7.00
ASB Yagi-Double stacked 6 element	12.70
ASA Yagi-Double stacked 370 to 430MC	29 .40

WESTINGHOUSE HYPERSIL TRANSFORMER

		_			
PRI-	115V.		60CY	3/4	KVA
SEC	#1 -	ď	240V		1.56A
SEC	#2 .		240V	-	1.56A
	WT		30 LE	S.	

\$14.50 EACH

Terms 20% cash with order, balance C. O. D. unless rated. All prices F.O.B our warehouse, Phila., Penna., subject to change without notice.

COAXIAL CONNECTORS



FULL LINE OF JAN APPROVED COAXIAL CONNECTORS IN STOCK

UHF	· N	BN	BN	C
UG-7 UG-12 UG-18 UG-19 UG-21 UG-21B UG-22 UG-22B	UG-23 UG UG-24 UG UG-27 UG UG-27A UG UG-30 UG UG-30 UG UG-34 UG	-57 UG-103 -58 UG-104 -83 UG-106 -85 UG-108 -86 UG-109 -87 UG-167	UG-175 UG-176 UG-185 MX-195 UG-197 UG-201 UG-206 UG-236	UG-254 UG-255 UG-260 UG-264 UG-274 UG-275 UG-290 UG-291 UG-306
M-358 M-359 M-359A M-360	MC-277 MC-320 PL-258 PL-259	PL-259A PL-274 PL-284 PL-293	PL-33 SO-23 SO-26 TM-2	39 34

TYPE "J" POTENTIOMETERS

D-163950 D-166132

ES-685696-5 ES-689172-1

\$1.25 each

49120 49121A

93-C 93-M

Resis.	Shaft	Resis.	Shaft	Resis.	Shaft
60	SS	5K	1/4"	50 K	3/8"
60	9/16"	5K	3/8"	50 K	1/2"
100	SS	5 K	1/2"	100K	SS
200	SS	10K	SS	100K	3/8"
500	SS	10K	3/8"	150K	1/2"
250	1/8"	10K	1/2"	200K	3/8"
500	5/16"	15K	SS	250 K	SS
500	1/2"	15K	1/2"	250K	3/4"
500	5/8"	20K	SS	250 K	3/8"
650	1/2"	25K	SS	500K	SS
1K	SS	25K	1/4"	500K	1/4"
2K	3/8"	30K	1 1/8"	500K	7/16*
2500	SS	40K	SS	1 Meg S	
4K	SS	50 K	SS	2.5 Mes	
5K	SS	50K	1/4"	5 Meg S	SS
DUAL	"JJ" P	OTENTI	OMETER	S\$1.9	5 each
50	SS	500	SS	1 Meg	SS
100	SS	1K	SS	2.5 Mes	SS
250	SS	2500	SS	5 Meg S	SS
330	SS	10K	SS	1K/25K	3/8"
Triple	"JJJ" Po	tentiomet	ers	SK-86	95 each.

SOUND POWERED TELEPHONES

U. S. NAVY TYPE M HEAD AND CHEST SETS U.S.I. A-260 W.E. D-173013 A.E. GL832BAO ANY TYPE—\$14.88 EACH TS-10 Type Handsets....\$8.92 ea.

F. W. BRIDGE SELENIUM RECTIFIERS

AC Volts Input 18	AC Volts Input - 40
DC Volts Out - 14.5	DC Volts Out - 34
1.3 Amps 3.85	0.6 Amps 4.60
2.4	1.2 5.95
6.6 7.75	3.2 8.95
13.0 12.75	6.0 15.50
17.5 15.75	9.0
26 22.75	12 26.95
39 35.50	18
52 38.50	25 42.50
70 49.50	36 55.50

130 VAC 1/2 WAVE STACKS 75MA \$.88 | 150MA \$ 1.30 | 250MA \$ 1.75 100MA 1.10 | 200MA 1.57 | 400MA 2.60

GENERATORS

• Eclipse-Pioneer type 716-3A (Navy Model NEA-3A)
Output—AC 115V 10.4A 800 to 1400cy. 1 φ; DC 30
Volts 60 Amps. Brand New... \$38.50
• Eclipse-Pioneer type 1235-1A. Output—30 Volts
DC 15 Amps. Brand New—Original Packing. \$15.50

THYPATPONS & IGNITRONS

11111	711/0113 G	10111111	0110
OA4G	FG-41	FG-271	722A
CIA	FG-57	393A	873
1C21	FG-67	394A	884
2A4G	FG-81A	GL-415	885
2B4	91	KU-610	1665
2D21	FG-95	KU-623	1904
3C31	FG-105	KU-628	2050
4C35	FG-166	KU-634	2051
C5B	FG-172	WL-65?	5550
C6J	FG-178	WL-672	5551
FG-17	RX233A	WL-677	5552
FG-33	FG-235A	WL-681	5557
			5540

TEST EQUIPMENT

• 1-222A Signal Generator \$79.50
 Vibrotest Mod. 218 Megger* \$45.00
• TN-IB/APR-I Tuning Unit\$95.00
C-D Quietone Filter Type iF-16 10/220 V AC/DC
20 Amps* \$9.00
• TS-127/U Freq. Meter w/spares\$69.50
TO 127/O Fred, Hieler Wyspares
• TS-143/CPN Oscilloscope * \$95.00
 Dumont 175A Oscilloscope* \$225.00
 LM-20 Frequency Meter* \$49.50
 Gen. Radio 757-PI Power Supply* \$27.00
• 1-130 A Signal Generator* \$70.00
TS-6/AP Frequency Meter* \$42.00
13-0/AF Flequency meter 342.00
 L & N_KS-7470 Null Volt Test Set* \$60.00
 A.W. Barber Labs. VM-25 VTVM* \$86.00
TS-10A/APN Delay Line Test Set\$45.00
• TS-19/APQ-5 Calibrator\$75.00
 REL W-1158 Frequency Meter 160-220 Mc.\$32.95
• CWI-60AAG Range Calibrator for ASB, ASE,
ACV and ACVC Dadays
ASV and ASVC Radars\$39.95
 CRV-14AAS Phantom Antenna for Transmitters
up to 400 MC\$11.75
3 CM. Pickup Horn Antenna\$9.95
Gen. Radio 736A Wave Analyzer\$600.00
All Items New Except Where Noted * (Exc. Used
Condition)

MISCELLANEOUS EQUIPMENT

Amperex IB98 Gamma Counter\$ 9,87
Powerstat 1226-115/230V input-0-270V out
@ 9 amp 37.00
G.E. 2CV2AI Servo Amplifier
Sperry A-3 Hydraulio Servos
EIMAC 35 TG Ionization Gauge 5.95
ATR Inverters 6V DC to 110 VAC 60 cy 75W 19.95
ID-6/APN-4 Indicator
R-7/APS-2 Receiver 49.50
R-78/APS-15 Receiver
SCR-522 Transceiver
RT-7/APN-1 Transceiver—less tubes 6.95
FL-8 1020 cycle filter
RM-29 remote control unit. 8.95
RM-14 remote control unit
RTA-1B 12/24 V dynamotor
BC-1206-CM2 Receiver
CY-230/MPG-1 Radar Console575.00
G.E. Type JP-1 portable current transformer. 32.50
ASB-4 Radar equip. Complete. 69.75
AN/APS-13 less tubes 12.95
T-9/APQ-2 less tubes
BC-645A complete
RCA AVR-15 Beacon Recvr. 15.50
TBY Trans-Recyr
Pioneer Type 800-1B Inverters—28VDC to 120V
800 cy 7 amp AC (used)
G.E. Inverter—28VDC to 120 VAC 800 cy
750VA 1 d
Navy SD-3 Radar complete \$1200.00
Navy DP-14 Direction Finder complete\$385.00
many Dr. 14 Direction Finder complete \$385.00

PULSE TRANSFORMERS

UIAH 9278	UIAH 9340
9280	9350
G.E. 68G828	AN/APN-4 Block Osc.
G.E. 68G-627	Phileo 352-7149
G.E. K-2469A	Phileo 352-7150
AN/APN-9 (901756-501)	Phileo 352-7071
AN/APN-9 (901756-502)	Phileo 352-7178

AN/APN-9 (901/56-502 AN/APN-9 (352-7250) AN/APN-9 (352-7251) Westinghouse 132-AW Westinghouse 232-AW2 Westinghouse 232-BW2

AN/APA-23 RECORDER

Sweeps any receiver through its tuning range and permanently records frequency and time of received signals on paper chart. Power input—(motor) 27V DC 1.5A, and (recorder) 80/15V AC 60-2600 cy 135W.
Originally designed to record pulse or sine-wave modulated signals received by AN-APR-1, AN/APR-5, BC-348, S-27. SX-28. BRAND NEW ... \$147.50

SPRAGUE PULSE NETWORKS

SYNCHROS

Size 1, 3, 5, 6, 7 and 8 generators, motors, control transformers, differential generators, and differential motors in stock. 6DG 6G 7DG 7G A B M 1F 5B 5CT 5D 5DG 5F 5G 5N X 2J1F1 2J1G1 2J1H1 C-44968-6 C-56701 C-69405-2 C-69406-1

SEND FOR COMPLETE LISTING

715-19

PHILA. 6, PA.

Telephones - MARKET 7-6771-2-3

LEADING SUPPLIER OF ELECTRONIC & AIRCRAFT EQUIPMENT

A C MOTORS

5071930 DELCO, 115 V., 60 Cy., 7000 RPM.
PRICE \$6.50 EA.
TELECHRON SYNCHRONOUS MOTOR, Type
B3, 110 V., 60 Cy., 4 W., 2 RPM.

TELECHRON SYNCHRONOUS MOTOR, Type BC, 110 V, 60 Cy., 6 W., 60 RPM. PRICE \$4.00 EA.

ASTERN AIR DEVICES, Type J33, Syncho-nous, 115 V., 400 Cy., 3 φ, 8000 RPM. PRICE \$15.00 EA

HAYDON TIMING MOTORS 110 V., 60 CY.

TYPE 1600, 2.2 W., 4/5 RPM. PRICE \$3.00 EA. TYPE 1600, 2.2 W., 1/240 RPM. PRICE \$3.00 EA. TYPE 1600, 2.3 W., 1 RPM. PRICE \$3.00 EA.

TYPE 1600, 2.2 W., 1 RPM. PRICE \$3.00 EA.

TYPE 1600, 2.3 W., 1 RPM. PRICE \$3.00 EA.

TYPE 1600, 2.2 W., 1-1/5 RPM.

PRICE \$3.00 EA.

TYPE 1600, 3.5 W., 1 RPM. With shift unit automatic engaging and disengaging shaft.

PRICE \$3.75 EA.

TYPE 1600, 2.2 W., 1/60 RPM.

PRICE \$3.00 EA.

SERVO MOTORS

REMOTE INDICATING COMPASSES 26 V., 400 CY.

D C MOTORS

DELCO TYPE 5069625 Constant Speed, 27 V.
D.C., 120 RPM. PRICE \$15.00 EA.
JOHN OSTER TYPE C-28P-1, 27 V., 0.7 Amp.,
7,000 RPM, 1/100 H. P. PRICE \$5.00 EA.
JAEGER WATCH CO. TYPE 44K-2 Contactor
Motor, 3 to 4.5 V. Makes one contact per
second. PRICE \$2.50 EA.
GENERAL ELECTRIC TYPE 5BA10AJ52C, 27 V.,
0.65 Amp., 14 oz. in torque, 145 RPM.
PRICE \$6.50 EA.
GENERAL ELECTRIC TYPE 5BA10AJ37, 27 V.,
0.5 amps., 8 oz. in. torque, 250 RPM.
PRICE \$6.50 EA.
GENERAL ELECTRIC TYPE 5BA10AJ37, 27 V.,
0.5 amps., 8 oz. in. torque, 250 RPM.

GENERAL ELECTRIC TYPE 5BA10AJ18D, 27 V., 0.7 Amps., 110 RPM, 1 oz. ft. torque.
PRICE \$6.50 EA.

BARBER-COLMAN CONTROL MOTOR, Type
AYLC 5091, 27 V., 0.7 Amps., 1 RPM. Contains 2 adj. limit switches. 500 in. lbs. torque.
WHITE RODGERS ELECTRIC CO., Type 6905
No. 3, 12 V., 1.3 Amps., 1½ RPM, torque 75 In. lbs.

RECTIFIER POWER SUPPLY

Max. speed 5000 RPM. PRICE \$14.00 EA.

GENERAL ELECTRIC TYPE 6RC146. Input 230 V., 60 Cy., 3 \(\phi\), at 28 V. D.C. Continuous duty. Size 46" high, 28" wide and 17.5" Adeep.

Max. speed 5000 RPM. PRICE \$14.00 EA.

GENERAL ELECTRIC TACHOMETER GENERA-TOR TYPE AN5531-1. Variable frequency, 115 V., 400 Cy., 3 \(\phi\). PRICE \$20.00 EA.

GENERAL ELECTRIC TYPE 6RC146. Input 230 V., 60 Cy., 3 \(\phi\). PRICE \$20.00 EA.

GENERAL ELECTRIC TACHOMETER GENERA-TOR TYPE AN5531-2. Variable frequency, 120 Continuous duty. Size 46" high, 28" wide and 17.5" Adeep.

Max. speed 5000 RPM. PRICE \$14.00 EA.

GENERAL ELECTRIC TACHOMETER GENERA-TOR TYPE AN5531-1. Variable frequency, 115 V., 400 Cy., 3 \(\phi\). PRICE \$20.00 EA.

HIDDICATOR, Part No. 656029, 115 V., 400 Cy., 3 \(\phi\). PRICE \$20.00 EA.

PRICE \$25.00 EA.

PRICE \$25.00 EA.

PRICE \$25.00 EA.

115 V., 400 Cy., 3 \(\phi\). PRICE \$20.00 EA.

PRICE \$20.00 EA.

PRICE \$20.00 EA.

PRICE \$20.00 EA.

PRICE \$25.00 EA.

PRICE \$25.00 EA.

PRICE \$15.00 EA.

INSTRUMENT

INVERTERS

WINCHARGER CORP. PU-16/AP, MG750. Input 24 V. D.C., 60 Amps. Output 115 V., 400 Cy., 1 φ, 6.5 Amps.

PRICE \$75.00 EA

HOLTZER CABOT TYPE 149 H, Input 24 V.
D.C. at 44 Amps., Output 26 V. at 250 V.A.,
400 Cy., and 115 V., 400 Cy., at 500 V.A.,
1 φ. PRICE \$55.00 EA.

PRICE \$30.00 EA.

PRICE \$30.00 EA.

PRICE \$3.50 EA.

2JIG1 CONTROL TRANSFORMER, 57.5/57.5

V., 400 Cy. PRICE \$3.50 EA.

PRICE \$3.50 EA.

PRICE \$3.50 EA.

PRICE \$3.00 EA.

PRICE \$3.00 EA.

PRICE \$30.00 EA.

PIONEER TYPE 12116-2-A. Input 24 V. D.C. 355DG DIFFERENTIAL GENERATOR, 90/90 V., at 5 Amps. Output 115 V., 400 Cy. 1 \$\phi\$ at 45 watts.

PRICE \$100.00 EA.

PRICE \$100.00 EA.

SGENERAL ELECTRIC TYPE 5D21NJ3A. Input 24 V. D.C. at 35 Amps. Output 115 V., 400 Cy.
PRICE \$50.00 EA.

W. E. KS-5950-L2 Size 5G, 115 V., 400 Cy.
PRICE \$10.00 EA.

Output 115 V., 400 Cy., 1 \$\phi\$ at 1.5 K.V.A.
PRICE \$47.50 EA.

D C ALNICO FIELD MOTORS

CK1, PIONEER, 2 φ, 400 Cy. PRICE \$10.00 EA.
CK2, PIONEER, 2 φ, 400 Cy. PRICE \$14.00 EA.
CK2, PIONEER, 2 φ, 400 Cy. With 40:1 reduction gear.
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Radio Receiver Chassis Complete except for 13 tubes. This chassis with standard 19" panel front contains the receiver for 493.5 MC complete with power supply and an additional low voltage power supply that originally supplied the keyer BC 770 as described below. 110 VAC 60 cycles is the primary voltage.

Five 10 mfd-600 VDC oil filled GE condensers are used as filters. Five stages of 49 MC IF's.

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BC 769 TRANSMITTER

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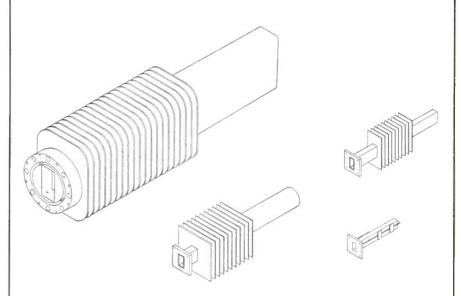
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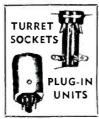
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3x0.1	600	Bot.	.40	2x1.0	600	(2) Side	.65
0.2	440AC	Side	.40	2.0	400	Side	.60
0.25	600	Top	.40	2.0	600	Bot.	.65
0.5	600	Bot.	.45	4.0	50	(1) Side	.65
0.5	600	Top	.48	50	25	(1) Side	.30
2x0.64	600	Side	.45	100	25	(2) Side	.65

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Mfd.	WVDC	Term	Ea.	Mfd.	WVDC	Term	Ea.
0.015	600	Top	.28	2x0.5	200	(2) Bot	.35
2x0.1	600	Bot.	.35	0.5	400	Bot.	.40
3x0.1	400	Bot.	.40	0.5	600	Bot.	.48
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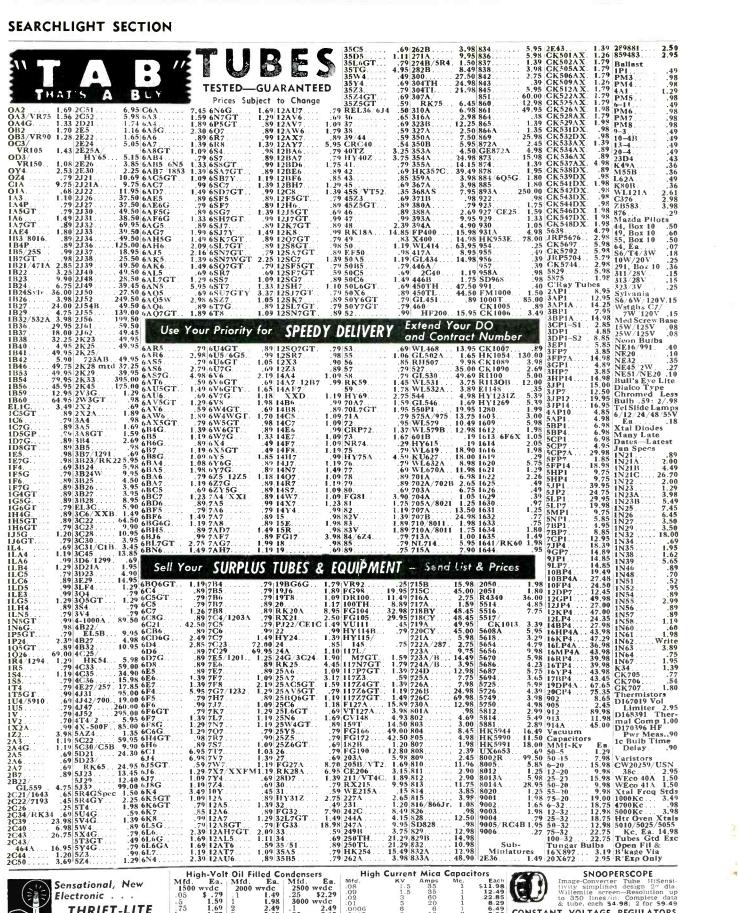
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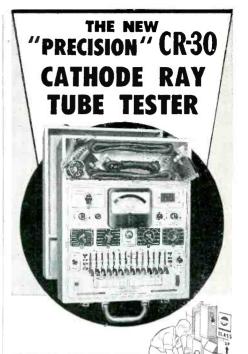
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TESTS ALL TV PICTURE TUBES (Magnetic and Electrostatic)

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GENERAL AND TECHNICAL SPECIFICATIONS

★ Tests All Modern Cathode Ray Tubes. Tests All CR Tube Elements: Not just a limited few. ★ Free-Point 14 Lever Element Selection System, independent of multiple base pin and floating element terminations, for Short, Leakage and Quality Tests.

Terminations, for Snort, Leakage and Quality lests.

* True Beam Current Test Circuit checks all CR Tubes with Electron gun in operation. It is the Electron Beam (and NOT total cathode emission) which traces the patern on the face of the CR tube. The significance of the above rests in the fact that Beam Current (and picture brightness) is primarily associated with the condition of the center of the cathode surface and not the overall cathode area.

★ Voltage Regulated, Bridge Type VTVM affords super-sensitive tube quality indications and positive check of low current anodes and deflection plates.

★ Micro-Line Voltage Adjustment Meter-monitored at filament supply

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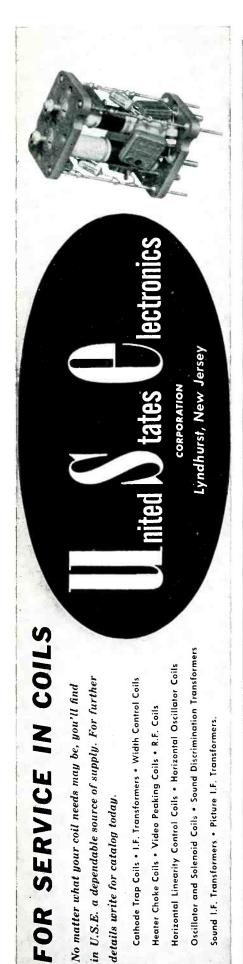
★ 45%" Full Vision Meter with scale-plate especially designed for CR tube testing requirements.

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Air Express Division. Rauland Corporation	177 43
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