

AUGUST • 1958 25 CENTS



**PF**  
PHOTOFACT

# REPORTER

FOR THE ELECTRONIC SERVICE INDUSTRY



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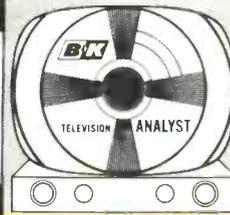
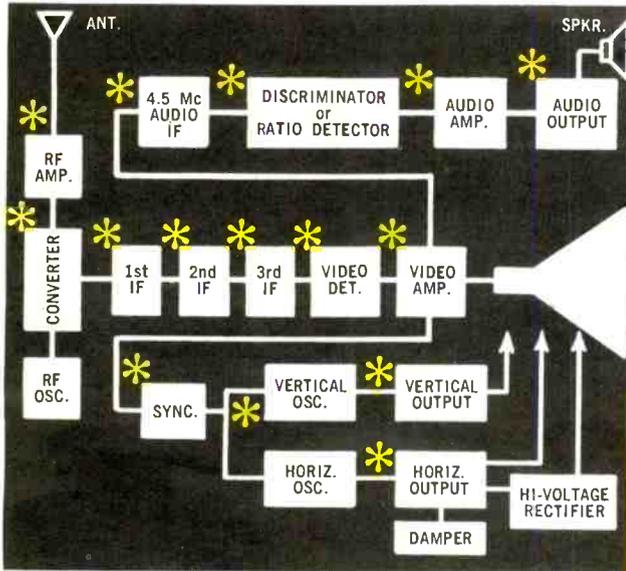


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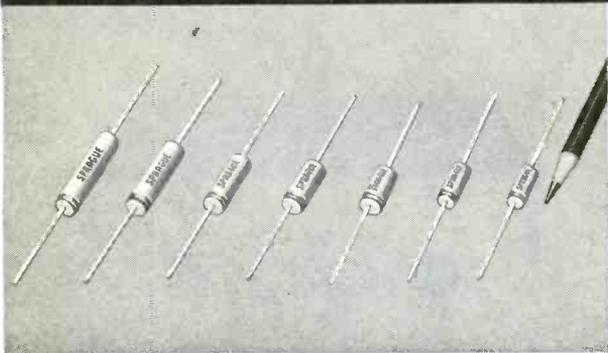
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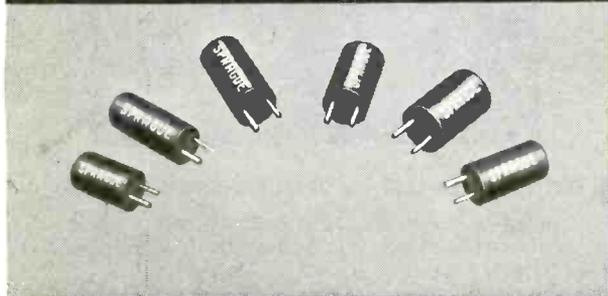
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*distributor for your copy of Sprague's Bulletins M-753 and M-757, or write Sprague Products Co., Distributors' Division of Sprague Electric Company, 105 Marshall Street, North Adams, Massachusetts.*

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## next month

### STARTING IN SEPTEMBER

A special fact-filled 8-page bonus section, comprised of two extra monthly departments containing data unduplicated anywhere else!

Four full pages of this special section will be devoted to photos and preliminary service data on the newest sets, providing you with all the advance data you need to service the set in the customer's home (particularly helpful if you've never seen the unit before).

The other four pages include trouble symptoms for various makes of TV receivers, complete with schematics of the circuits involved and explanations of the reasons for improper operation, plus what can be done to restore normal operation.

This new bonus section doesn't replace any of PF REPORTER'S regular departments or features—you'll be getting them all — plus this 8-page editorial section (containing no ads) each and every month starting in September.

VOLUME 8, No. 8



AUGUST, 1958

# PF REPORTER

FOR THE ELECTRONIC SERVICE INDUSTRY

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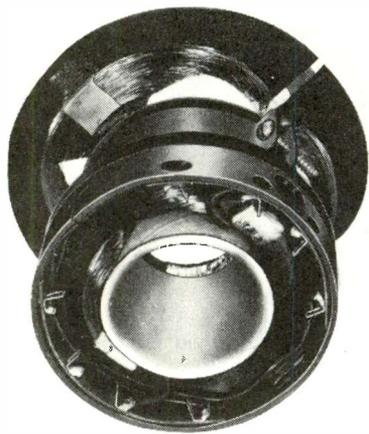
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**TRANSFORMER NEWS** FROM TRIAD



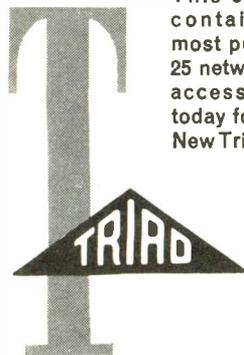
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*Letters to the*  
**EDITOR**

In the May "Troubleshooter" column, Mr. Norris Asche mentions installing a *retrace suppression* circuit in a set. This is new to me, and I would like to know more about it. Is it a standard circuit or must it be custom-designed for each application?

ERNEST E. BIEN

Sharon Hill, Pa.

*Perhaps you know this circuit by the name of retrace-blanking network. Turn to Jesse Dines' article in this issue for a complete explanation.—Ed.*

Dear Editor:

I am building a preamplifier like the one described in your June, 1955 issue. The article refers to a previous issue, which I would like to acquire if possible.

In the diagram given on page 25 of the June, 1955 issue, one side of the equalization selector switch is not connected and there is some doubt in my mind as to where I should connect C8, C9 and C10.

On the same schematic, could you recommend a value for R38 that would equalize a General Electric VR11 cartridge for proper RIAA response?

LOUIS ROSSI

Oakland, Calif.

*The preamplifier described in June, 1955 is a modernized version of one first introduced in our July-August, 1952 issue. The original unit had a 6X4 rectifier and AC-powered tube heaters, but otherwise was practically the same as the later design.*

*The open contact on the equalization switch should connect to the junction of R16 and R20. The value of R38 for equalization of the VR11 cartridge is 6.8K ohms.—Ed.*

Dear Editor:

There is a standard procedure which most top TV men use for approaching any service problem. It's hard to diagram, but it's there and makes them professionals. Can you emphasize this approach in some of your articles? Even the worst problem is simple—if you don't panic.

KEAT N. THOMPSON

Gainesville, Fla.

*What you have in mind is evidently the same thing Art Margolis was talking about in "A System for Starting on Dogs" (February, 1958 issue) or which Stan Prentiss outlines in this issue. To give it a fancy name, it's an analytical approach. The expert keeps track of every step in his work and has a definite purpose for every action he takes. This method keeps him from going around in circles when trying to lick a tough problem. Most of*

*the material in PF REPORTER is intended to help the reader in using this approach.—Ed.*

Dear Editor:

In your June "Letters" column, you advise that reprints of the "Tube Substitution Guide" presented in "Quicker Servicing" are available on request. Would appreciate your sending me a copy.

I have a TV technician friend in Sydney, Australia who I'm sure would also like a copy. May I have an extra to send to him.

G. R. MILLER

San Carlos, Calif.

*Okay, Miller, two copies are on their way—yours is the best excuse yet for getting an "extra."—Ed.*

Dear Editor:

Thank you for reinstating "Audio Facts." Let's have more articles on "old-fashioned" audio, now that it has been rejuvenated by hi-fi. Incidentally, what can be "hi-fi" about lipstick? Such a brand name is now on the market. Sounds as if everyone is jumping on the bandwagon!

WILLIAM E. STEPHENS

Jackson, Mich.

*It might not be "true hi-fi," but it's loud.—Ed.*

Dear Editor:

I've been a subscriber for over a year now, and learn something helpful from every issue. My favorites are "Notes on Test Equipment" and "The Troubleshooter."

There is something else I would like to have you include, however, and that is a preview of what is new in each year's designs. It would give us an opportunity to know about new sets before we are called upon to service them.

DAVID R. LYBROOK

Bakersfield, Calif.

*Perhaps you saw "Previews of 1958 TV Sets" in our November, 1957 issue. Starting next month, there'll be an 8-page BONUS section in EVERY issue—including 4 full pages of "Previews" which will provide you with preliminary data to help you service new sets when you first encounter them in the customer's home. Okay, Dave?—Ed.*

Dear Editor:

Jesse Dines' article "Analyzing the Vertical Circuit" in your June issue is one of the most comprehensive I have read in many a day. It is written with the serviceman in mind, but even an "egg-head" could learn much from it. Let's have more articles by Jesse.

EUGENE P. CURTIS

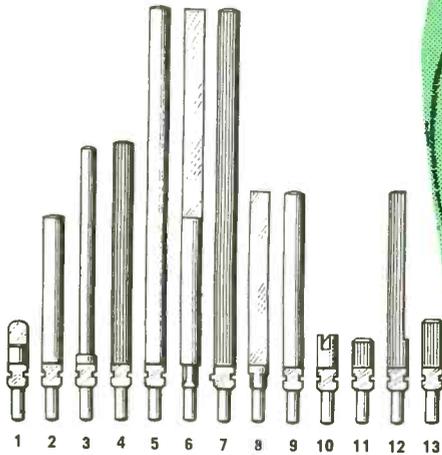
Portland, Ore.

*You guys really flooded the mails in response to this article. See page 10 of this issue for another Dines' original.—Ed.*

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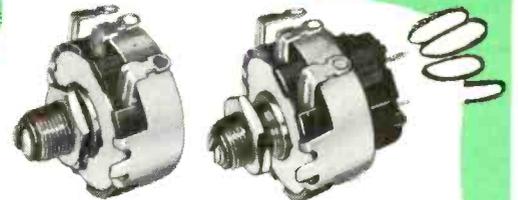


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**D. GIANT TUBE CARTON.** Eye-arresting... indicates the quality components you install and sell.

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All your display needs are met by smart new G-E designs—from an eye-catching, easy-to-apply door or window decal, to metal and illuminated outdoor and indoor signs... plus colorful utility items like the electric clock and thermometer.

These brand-new signs and displays are waiting for you now at your General Electric tube distributor's. Ask him how to obtain them! *Distributor Sales, Electronic Components Division, General Electric Company, Owensboro, Kentucky.*

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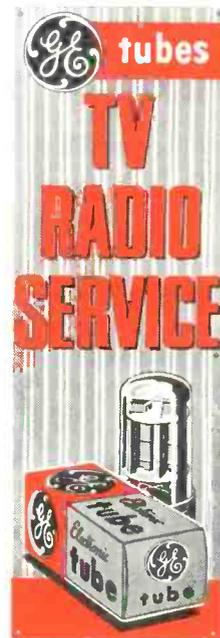
**COUNTER-WALL-WINDOW SIGN.** Illuminated. Use it anywhere in your shop! Wrap-around cover has the attractive display on both sides. 15" by 11½" by 5".



**OUTDOOR-INDOOR THERMOMETER.** Tells temperature accurately. Every viewer will see your advertising message. Replaceable front. 12" diameter.



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**METAL TACK-ON SIGN.** Narrow, smart dimensions—14" by 42". Fasten against building, or mount as a free-standing display.

# across the BENCH

## Old TV's Never Die by Stan Prentiss

While these original words referred to "old soldiers," they apply equally to old television receivers, which also "fade away." Cathode-ray tubes grow weak, capacitors and resistors age, tubes accumulate interelement leakage and lose emission, and the wrinkles of age multiply. But unlike "old soldiers," a good television receiver can be given new life, so that it will perform as good as (or even better than) it did when first installed.

### There is a Difference

Of course, there is a difference among "old" receivers. Regardless of the competence of the manufacturer, there seem to be good years and poor years for his products, depending on a multiplicity of factors. Engineering, research, materials, labor, etc., all play a respective part, and the result may not be known until the product has proved itself with the passage of time. One of these is used in the following example of what can be done with an old set.

### Admiral 36X36

The sensitivity of this receiver is nothing to brag about by today's standards, but its reliability for day-in, day-out operation is hard to beat.

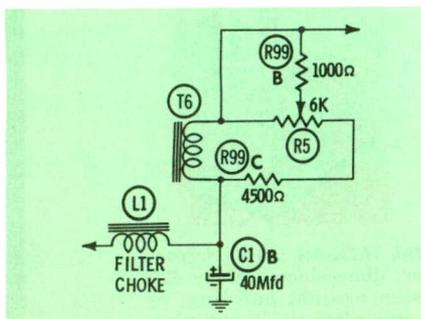


Fig. 1. Focus control in Admiral 36X36 is used to vary resistance across coil.

It was introduced about 7 years ago (1951) as a composite of, and improvement on, several predecessors which were themselves durable workmen. 1951 was a little early for employment of the cascode tuner; however, a good Standard Coil pentode type was used. Three stagger-tuned IF stages support the tuner, and the audio section (with a big, beautiful 6V6 providing 5 watts of potential output) is supplied with signal from the IN64 video detector diode. An old reliable 6AC7 (you'll find plenty of originals still in these sets) provides a positively-polarized video signal (sync positive) to the cathode of the picture tube.

The vertical sweep generator is of the blocking-oscillator type with a 6S4 tube and circuit of conventional design. The horizontal section is composed of the familiar phase-detector AFC system, where a pulse-developed sawtooth from the high voltage transformer is compared with pulses from the sync chain to generate a control voltage for the input grid of the horizontal multivibrator. A 200-volt positive spike from the same high voltage transformer terminal drives the AGC keyer tube, which regulates the bias for the RF and the first two IF amplifiers. The sync chain is standard and requires no particular explanation other than to remind the technician that for waveform viewing in the sync separator, clipper, and inverter stages, oscilloscope sweep selections correspond to either the horizontal or vertical scanning frequencies, depending upon the objective sought.

### The Trouble

Recently, I had occasion to examine one of these sets. Upon arrival

in the customer's home, I detected a sputtering noise, which obviously came from the back of the receiver. There was neither sound nor picture and the odor reaching my nostrils told me I'd better turn off the power. Unfastening the rear cover and re-connecting the AC line cord revealed a sparking low-voltage rectifier tube and a curl of smoke ascending from the front of the chassis.

"We have been told," said the customer, "that the main transformer is shorted and that repairs will cost at least \$70. This is an old set, and even though it's always performed well, we just don't want to spend that much money to have it repaired."

This is sort of a routine attitude among many owners of older receivers. Either they have already spent a considerable amount of money on the set, or they are apprehensive about spending more than its apparent value justifies. In this instance, however, there obviously remained an element of doubt. "We have been told," the woman said, and along with the statement was a rather evident desire to have the old receiver back in operating condition if such was possible at a reasonable fee. Here were all the psychological elements necessary to promote a profitable repair plus an opportunity to magnify an already good reputation with a new client (yes, client—"one who employs the services of any professional man"). In short, a chance for more business!

### What Might Have Been Done

Under this set of circumstances there are two usual alternatives: either you courteously remove the receiver from its cabinet and make some preliminary tests to try and determine the extent of the necessary repairs, or you agree with the supplied diagnosis and offer to sell the customer a new receiver.

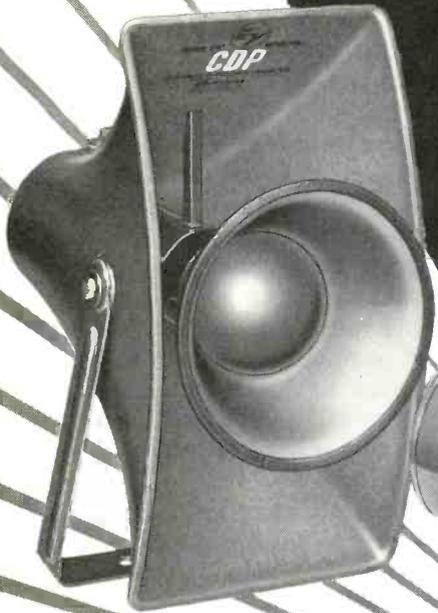
That last alternative would have backfired immediately. The customer wanted both assurance and insurance. The first thing that had to be done was to convince them of my ability as a television technician; the second was to repair the receiver at a logical, if not reasonable, price. To the customers, in this instance, a reasonable price would naturally be anything *under* the previously

• Please turn to page 52

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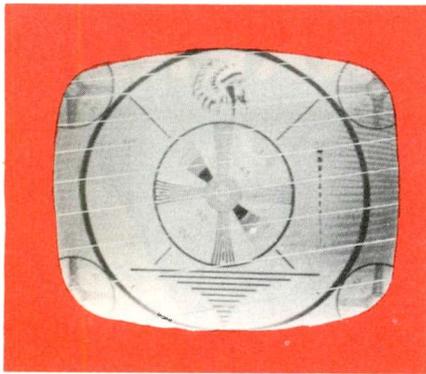


Fig. 1. Retrace lines on a picture are very disturbing, but can be eliminated.

# vertical RETRACE-LINE elimination

By Jesse Dines

Bring Those Old Sets Up To Date  
By Installing Blanking Circuits.

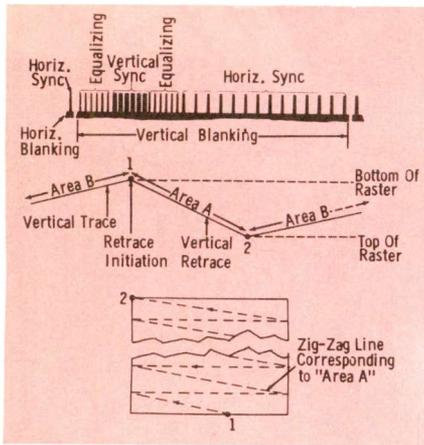


Fig. 2. A comparison of signals and beam action during vertical retrace.

Occasionally, you may come across a receiver whose raster has visible vertical retrace lines such as pictured in Fig. 1. These retrace lines are annoying to the viewer, especially when they can be seen in the picture background. In most cases, the receiver is operating in accordance with the manufacturer's specifications; yet, if you can modify the design to obtain complete retrace blanking, your customer will most certainly be grateful.

## Cause of Retrace Lines

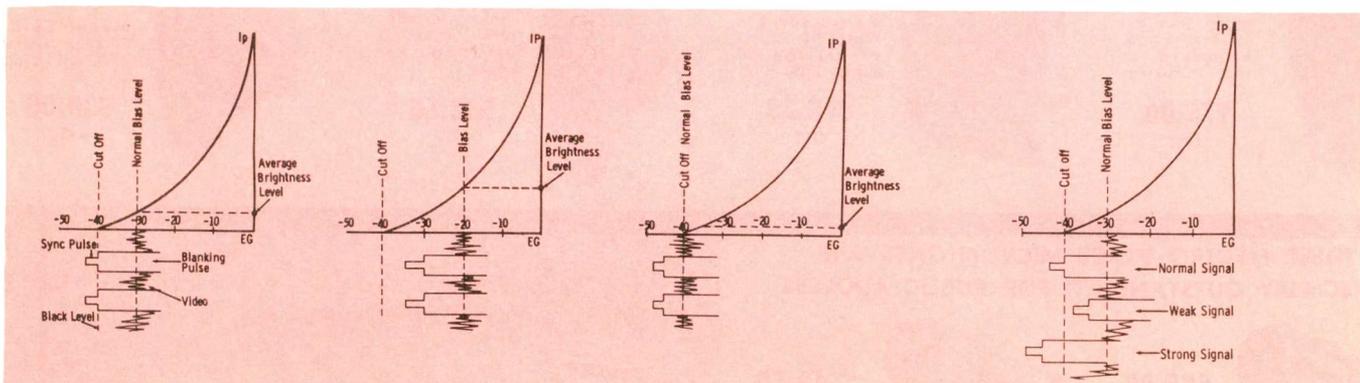
Before discussing retrace elimination circuits, let's first see what causes retrace lines. Examine the composite video signal shown in Fig. 2A, in which the various components are labeled. A waveform of the current which produces vertical raster deflection is shown in Fig. 2B. The vertical trace portion of the sweep current causes the picture tube beam to be deflected from top to bottom in less than 1/60 of a second, completing one field. The vertical retrace portion, which is initiated when the vertical oscillator triggers, deflects the electron beam back to the top in something like 1200/1400 microseconds (depending on the receiver). This time is designated as "Area A" in Figs. 2B and 2C. Fig. 2C shows how the retrace lines form a zig-zag pattern because horizontal scanning continues during vertical retrace.

Now let's change our trend of thought slightly and see what happens under conditions of normal and abnormal brightness control settings. Typical Eg- $I_p$  curves for a picture tube and changes in bias caused by video signal excursions are shown in

the illustrations of Fig. 3. Operating levels for various control settings are shown so that the picture tube has (A) normal bias, (B) too little bias, and (C) too much bias. The bias normally developed is effectively the average AC potential of the video signal, and average beam current corresponds to this DC level.

In Fig. 3A, the brightness control is set so that picture tube bias is an arbitrary 30 volts. Note that the 40-volt cutoff point occurs at the black level (at sync tips). In this case, blanking pulses effect picture tube cutoff and retrace lines are not visible.

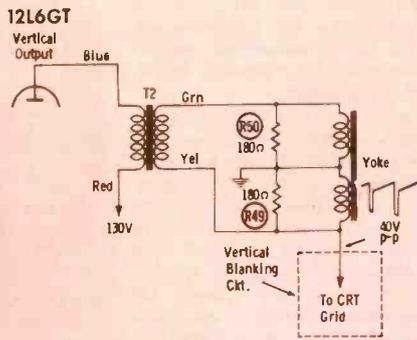
In Fig. 3B, insufficient bias is produced by the brightness control setting; thus, sync and blanking pulses fall short of cutoff, black portions of the video signal will appear as grays. All portions of the image will be brighter than normal, and may even have a "washed out" appearance. Inasmuch as the picture tube is not cut off during vertical retrace, the undesired lines will appear. At this point, perhaps you can see that vertical blanking would permit good picture viewing with a higher average brightness level by preventing



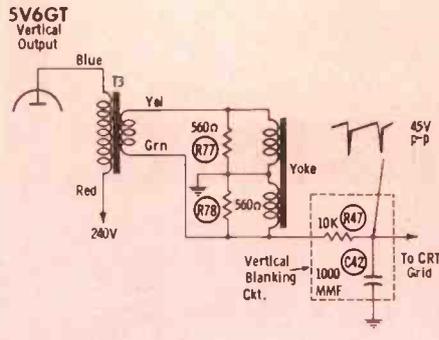
(A) Normal bias. (B) Too little bias. (C) Too much bias.

Fig. 3. Picture tube Eg  $I_p$  curves showing how bias affects image reproduction.

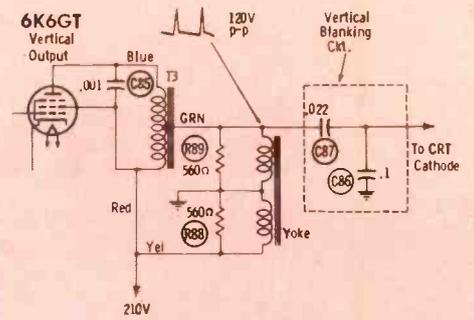
Fig. 4. Picture tube Eg  $I_p$  curve showing effects of changes in video signal level.



(A) Truetime Model 2D3770A



(B) Emerson Chassis 120257-D



(C) RCA Chassis KCS88J

Fig. 6. Typical vertical blanking circuits incorporated in modern receivers.

retrace lines from being seen.

In a similar analogy, a signal with insufficient amplitude (due to a weak signal or a misadjusted contrast control) may also result in visible retrace lines (see Fig. 4). Note that the effects of a normal signal, weak signal, and strong signal are similar to those indicated in Fig. 3A, B, and C, respectively.

### Obtaining the Spike

To insure vertical blanking, a portion of the vertical sweep energy can be shaped and fed to either the grid or cathode of the picture tube. If pulses are fed to the cathode, they must be positive in polarity; if fed to the grid, they must be negative.

In either case, the picture tube will be biased to or beyond cutoff during the proper interval.

The spike acts in conjunction with the vertical blanking portion of the composite video signal to drive the

picture tube into cutoff. To avoid adverse loading of the video amplifier, it is usually a good idea to feed the spike to the undriven picture tube element. Thus, if the video amplifier feeds the grid, connect the spike voltage to the picture tube cathode and vice versa.

Fig. 5 represents typical voltage waveforms found in the vertical sweep circuit. It can be seen that a negative-going spike is present in the plate circuit of the vertical oscillator tube and one of positive polarity at the plate of the output stage. Note that a negative spike appears across resistor R1 and a sawtooth across capacitor C1 in the sawtooth-forming network of the vertical oscillator; hence, a blanking signal from the junction of R1-C1 would be applied to the picture tube grid. Note, too, from Fig. 5, that the spike voltage obtained from the output

• Please turn to page 58

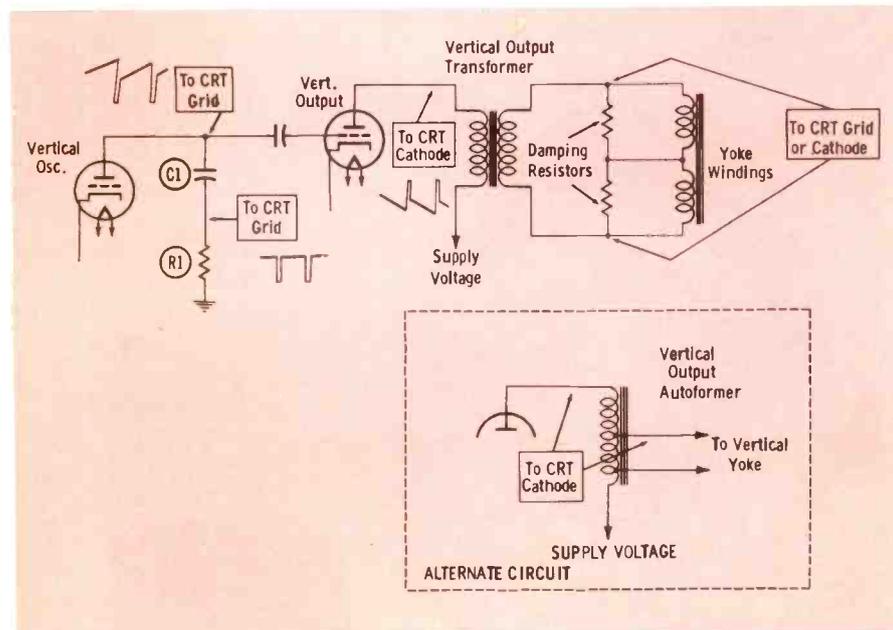
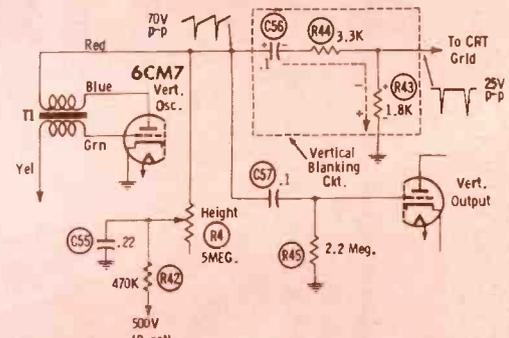
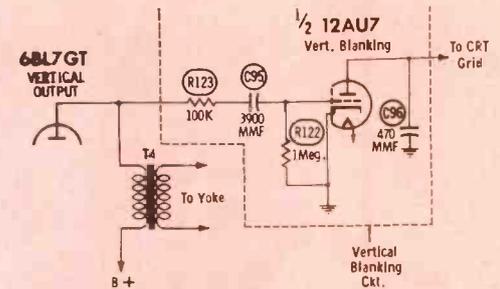


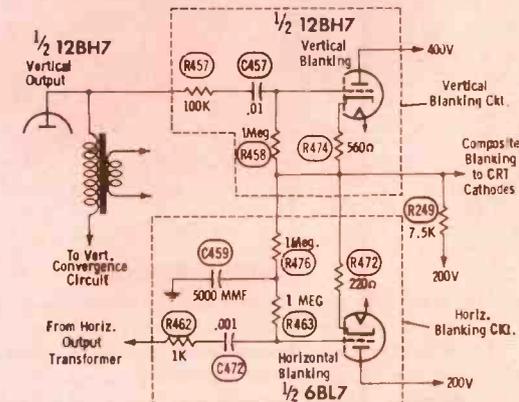
Fig. 5. Partial schematic of vertical section shows blanking take-off points.



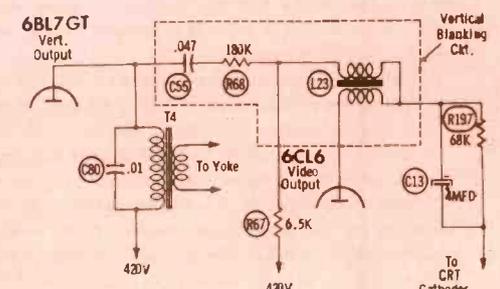
(D) Airline Model GRX4020A



(E) GE Model 24C101



(F) GE Model 15CL100

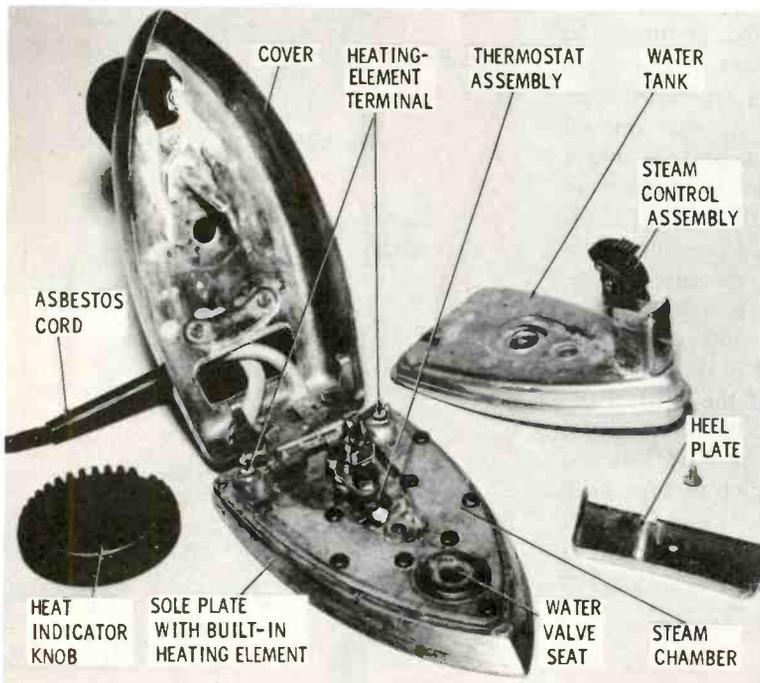


(G) Raytheon Chassis 21CT1C

# SERVICING SMALL APPLIANCES

If you can repair radio and television receivers, you can also fix small appliances, for you already have the basic essentials necessary—electronic background, tools, and test equipment.

Naturally, you should feel your way slowly at first—but when you see how profitable it can be (perhaps helping you through a temporary slump), you may want to go ahead and invest in special test instruments such as surface- and water-temperature meters, a tachometer to check the speeds of motor-operated devices, etc. One problem, of course, is getting replacement parts and adequate service literature. You might plan ahead, however, and contact as many distributors and appliance manufacturers as possible before a particular need arises. The items pictured and described here are just a fraction of the many appliances found in the average home. And to help you get started in repairing them, we've provided you with pertinent data and servicing tips.



## Irons

The electrical circuit of an iron is relatively simple, consisting basically of only a heating element attached to an AC cord. Automatic and steam irons, of course, have certain features such as variable heat controls, but these are not at all complex. Most of the troubles encountered will be either shorts or opens. Little need be said about the continuity tests for locating these sort of troubles.

You will probably find that the toughest job is taking the iron apart. For this reason, obtain all the disassembly information you can from manufacturers and their distributors.

Don't waste valuable time repairing parts such as open heating elements, defective thermostats, or leaky water tanks, which are more easily replaced.

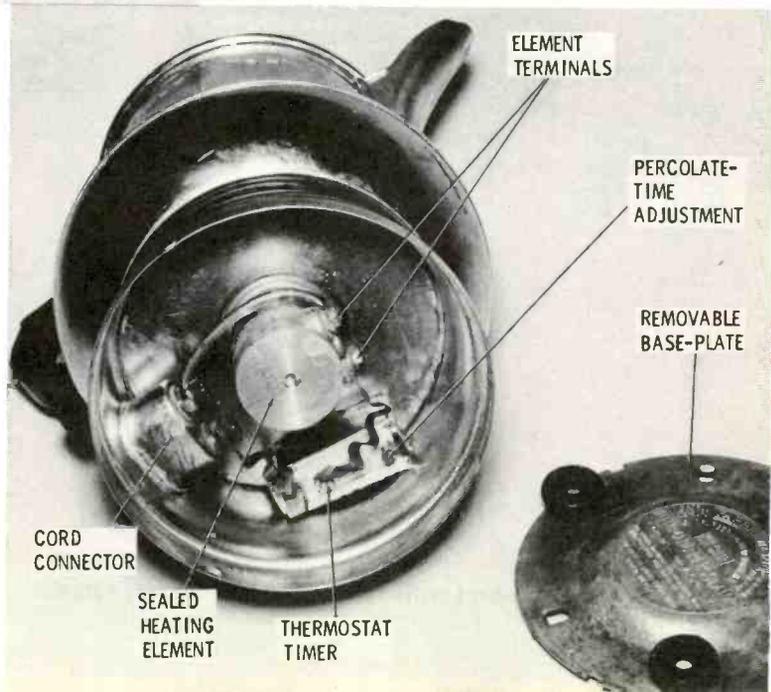
If ordinary tap water has been used in a steam iron, its steam passages, valve, and port holes may become congested with lime and other chemical deposits. Before disassembling the iron, try cleaning it with a small amount of white vinegar. Let the iron "steam" until the vinegar starts to boil; remove the plug and let it sit for twenty minutes; then flush with pure distilled water.

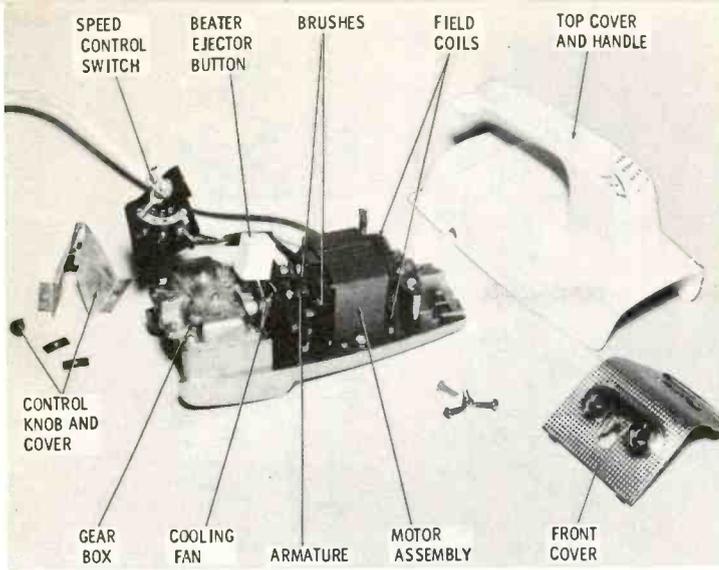
## Coffee Makers

Normally, you will encounter two general types—the percolator and the brewer. Either, however, may come in automatic or non-automatic designs. When servicing non-automatic percolators, check the heat fuse which protects the unit against damage when the pot boils dry. Always test for shorts before replacing this fuse. Inside the percolator is a stem with a coffee basket at the top, and a hood or valve at the base. Check to see that the valve is clean and that it seats firmly over the heating element.

Automatic percolators, such as shown, will feature a thermostatically-controlled timing unit of some sort: If this proves to be at fault, replace the entire assembly.

In the automatic brewer, more than one heating element may be employed. Heat controls automatically switch operation from a high-heat to a low-heat element. The first causes the water to boil, and steam pressure forces it up into the brewing basket. When pressure drops, the brew returns to the bottom section, and the second element keeps it at serving temperature. Adjustment of these controls should be in accordance with the manufacturer's recommendations.



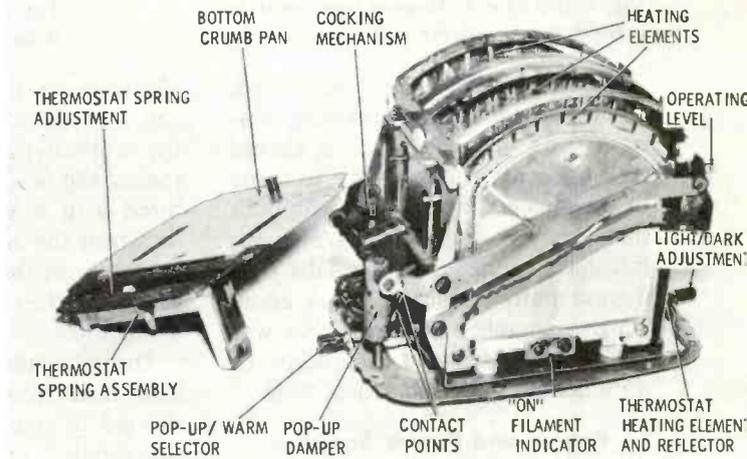


### Toasters

All toasters operate on the same basic principle wherein a thermostat opens a switch when a certain temperature is reached. To accomplish this, the unit may employ a clock-timer, a direct thermostat, an adjustable timer, or a combination timer and thermostat.

Removing the outer covering of a toaster may be a little tricky—but with experience, you'll find it no more difficult than disassembling a TV high-voltage cage. Remember, however, not to go any further than necessary when taking a unit apart. The heating elements are relatively delicate, so go easy when probing around for trouble.

The most common troubles found in toasters are damaged carriages or cocking mechanisms, dirty or gummed-up parts, and misadjusted thermostats. In some cases, the heating elements may be broken or burned in two. If at all possible, replace the element rather than try to repair it. In older units for which replacements are hard or impossible to obtain, however, use a mending sleeve to repair the open element.



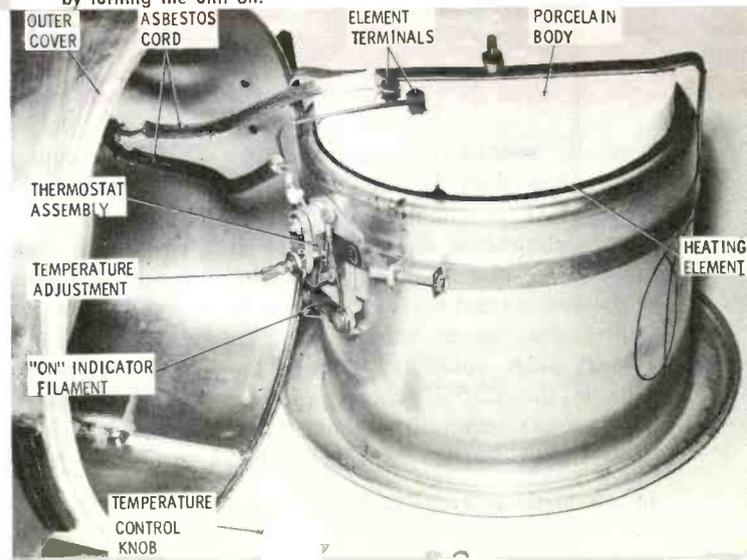
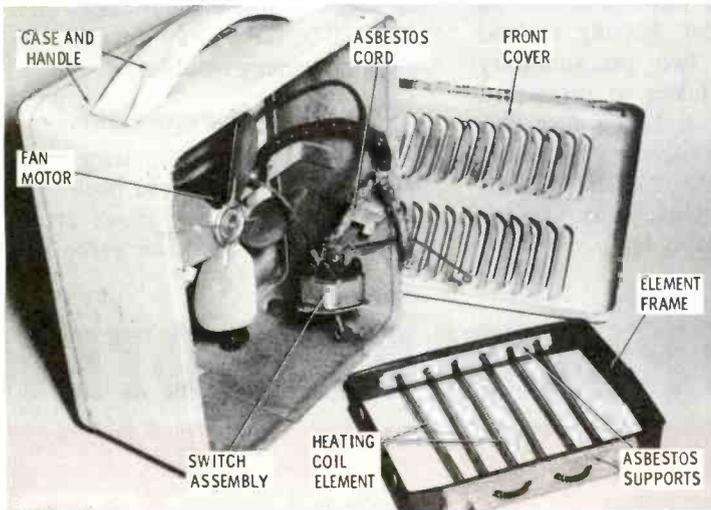
### Space Heaters

Small space heaters are nothing more than a heat-producing grid with a fan or reflector to direct the air or heat waves. Aside from fan trouble, it is not unusual to find a faulty switch or a burned connection. The switch may regulate current through the heating element and/or control the speed of the fan motor.

Make sure all electrical connections are tight. At high currents such as found in heaters, a loose contact will usually develop enough excessive heat to completely burn off a terminal or its mounting board.

As with any other appliance, the test procedure should include a thorough inspection of the line cord. Since the cord undergoes more abuse than any other component, it naturally becomes a common source of trouble. If you question its current-carrying capabilities for any reason, replace it.

If you are unable to obtain a new heating element of the resistive-wire type for a particular heater, a temporary repair of an open can be made by placing the wire ends together, putting a small amount of borax on the joint, and welding the junction by turning the unit on.



### Cookers

This group of appliances includes roasters, casseroles, broilers, deep fryers, hot plates, etc. When a unit of this type will not heat, check first for continuity through the power cord and heating element.

If the cooker has a thermostat, its operation and adjustment should be checked according to manufacturer's specifications. Always be sure that the heat-conducting portion of the thermostat is making good contact with the inner heating surfaces of the cooker. Also, check to see that the heating element is correctly positioned, a factor which has a definite effect on thermostat operation.

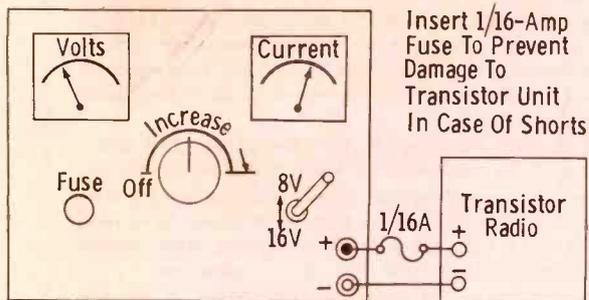
Most small appliances will have either painted, chrome, or porcelain-enamel finishes; therefore, take care in handling them by padding your workbench and using a cloth or piece of cardboard between your tools and such vulnerable surfaces.

### Mixers

In nearly all domestic food mixers and blenders, a series-type commutator motor is used. Check the motor for overheating, irregular speed, excessive arcing at commutator, loose bearings, or armature binding. If the motor chatters, chances are that the bearings are badly worn and should be replaced. One can check the speed control and field coils for opens or shorts by making a simple ohmmeter test.

When checking the gear box for worn or damaged gears, always be on the lookout for oil leaks. If you disassemble the motor and gear case, it's a good idea to replace all oil retainers to insure a leak-proof seal. If the mixer motor refuses to run and the power cord is okay, check the switch assembly, solder connections and leads, motor brushes, field coil and armature.

Use factory-replacement parts whenever possible. If you estimate that the cost of parts and labor will exceed the price of a new appliance, consult the customer before going ahead with the repair.



Insert 1/16-Amp Fuse To Prevent Damage To Transistor Unit In Case Of Shorts

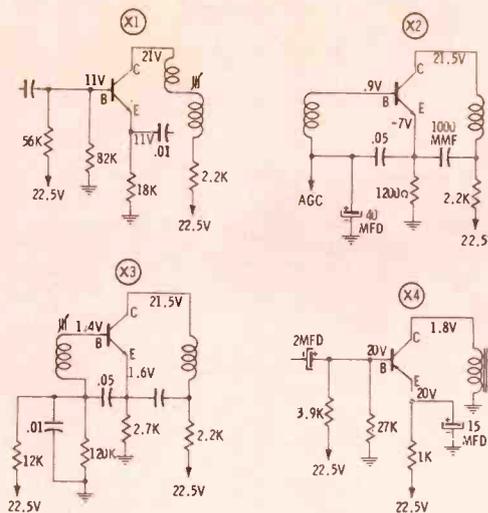


Fig. 2. Partial schematic shows voltage-divider networks and dropping resistors which can be used to locate test points.

Fig. 1. Use of a 1/16-amp fuse helps to protect the transistor radio under test.

Previously, we have said much about basic transistor servicing concepts, and many technicians should now be ready for exposure to some of the more advanced techniques that can be employed. We have, through experience in our labs with a great many transistor radios, come across a number of points that will enable the advanced technician to do a better repair job in less time.

### Power and Power Supplies

One of the first obstacles you will encounter is how to power the transistor radio during necessary test procedures. The newer models of battery eliminators, designed for servicing transistorized auto radios, were checked as possible power sources. A scope analysis of output current ripple content with the drain

obtained on the 8-volt range. Because of the low voltage ratings of the electrolytics used in transistor radios, the output voltage was monitored with a VOM and a red line drawn at the rated voltage point on the meter of the power supply. Current was then monitored with the VOM.

During a subsequent test, the positive lead from the power supply shorted to one lead of a miniature electrolytic, ruining the capacitor and blowing a 1" section of foil on the printed board. Having learned the hard way, two precautionary measures were taken to prevent a recurrence: (1) a 1/16a fuse was inserted in the positive lead from the power supply (Fig. 1), and (2) insulated alligator clips were installed on the power supply leads.

check the battery current drain? First of all you tune off station, then adjust the volume control to maximum and note the reading on the current meter. If current is much higher than normal (more than 10-15%), the set probably has a leaky capacitor or a transistor that is biased improperly. If current is too low, then one stage isn't working due to an open capacitor, open resistor, broken connection, high value resistor, open transformer, etc. As you can see, a simple current check puts you well on the way toward the solution to many troubles.

### Voltage Measurements

Various articles on servicing transistorized equipment have been very consistent in stating, "Don't check resistances in a transistor radio un-

# ADVANCED TROUBLESHOOTING

# TRANSISTOR

normally required by a transistor radio (less than 10 ma) revealed that, for all practical purposes, the battery eliminator was as good as a battery. We also found that because of the low current drain of transistor radios, the output voltage on the 16-volt range could be adjusted to provide the 22.5V (even more in some cases) required by some transistor radios. Further, 9 volts can be

This setup, although a workable one, is rather bulky and presents certain dangers to the equipment under test. A power supply designed for use with transistor radios is much smaller and eliminates these dangers. In addition, accurate voltage and current meters are often provided on such instruments, which makes checking for proper current drain much easier. How do you

less you remove the transistor from the circuit under test." This author suggests you take it a step further; forget about resistance and check voltage instead. After all, a resistance check is made under dynamic conditions and will therefore provide more data in less time. Because the voltages employed in a transistor radio are necessarily low, and minute variations are often very signifi-

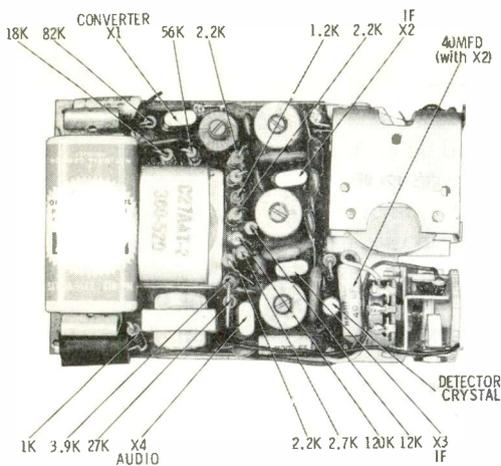


Fig. 3. Sample layout of radio showing locations of transistors, helpful guides.

cant, the meter employed must be quite accurate. A recently-calibrated VTVM fills the bill very nicely.

Before you can measure voltages, you have to locate the desired check points on the printed wiring board. Rather than spend a lot of time trying to trace the crazy maze of silver ribbons, consult the schematic and the pictorial component-identification sections of the service literature and systematically locate the chosen test points. (ED. NOTE: Beginning with Set No. 414, PHOTOFACT folders will identify various circuit connections by keying them to the schematic.) In Fig. 2, we have shown a partial schematic of the four stages employed in a typical transistor radio. In each stage, the resistors (easiest components to identify) can be used as "land-

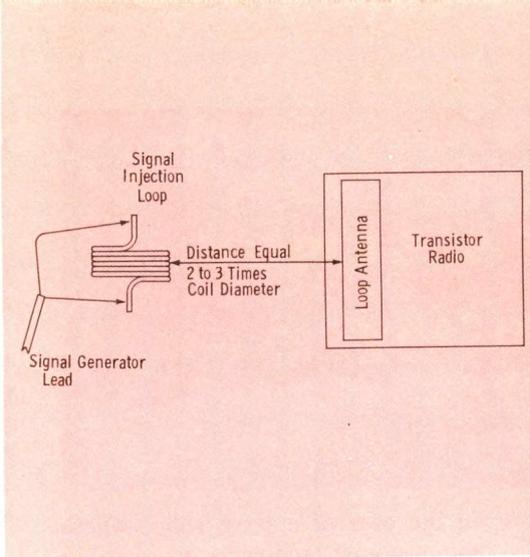


Fig. 4. How to radiate signal into the loop antenna of a transistor radio.

Once you have determined where to make the checks, it is a simple matter to check the voltages present at the elements of the transistors. A rapid check of a transistor circuit consists of measurements at the base and emitter. These voltages should be nearly equal as indicated in Fig. 2. Any significant difference between the voltages on these elements is an indication of some malfunction in that stage. The readings given in the service literature provide an accurate reference point in each case.

#### Alignment

The alignment of a transistor radio is not a hit or miss proposition; it is an exact science. The instructions given in service literature provide an accurate guide for the actual procedure; however, you must have

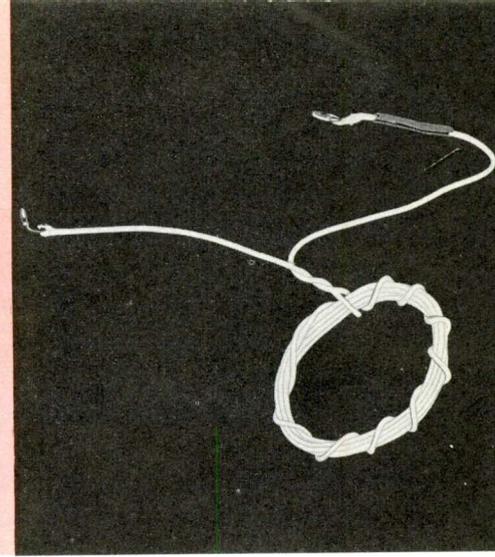


Fig. 5. Sample of signal injection coil made from 15 turns of number 26 wire.

not detune or unnecessarily load the input circuit. A commonly recommended procedure for signal injection involves radiation of the signal into the loop antenna of the receiver. This should be done as illustrated by the drawing in Fig. 4. Notice that the injection coil is positioned at right angles to the loop, and at a distance of about two or three coil diameters. The exact distance must be determined by trial and error since it is affected by radiation characteristics of the coil, signal generator output voltage, and sensitivity of the loop antenna circuit. A good workable injection coil (Fig. 5) is comprised of 15 turns of No. 26 insulated wire wound into a coil 1½" in diameter. The last turn should entwine the others to hold them together. (Always connect the ground lead of the generator to this end of the coil.) In lab tests we found that getting the injection coil too near the loop resulted in an apparent detuning of the circuit whenever a hand was brought close enough to the trimmers to make adjustments. Moving the injection coil farther away from the loop minimized this effect. Along with signal injection, we should consider the output indicator; a VTVM connected across the end terminals of the volume control is the most satisfactory.

#### Oscillator Tracking

In an examination of several radios, two differences between transistor oscillator circuits and their vacuum-tube counterparts were noted. One is that an adjustable oscillator coil is always employed, and the other is that the sections of the

• Please turn to page 55

# TECHNIQUES FOR RADIOS

marks." If you study the parts layout (a sample showing transistor locations is shown in Fig. 3) and positively locate stages 1, 2, 3, and 4, you can associate the resistors with the adjacent transistor and identify the desired test points. Make yourself a little chart if you can't remember them. A few minutes spent in this manner can save much time in the over-all job.

a great deal of personal knowledge about circuit operating theories and designs if best tracking and sensitivity are to be achieved.

#### Signal Injection

One of the first problems that arises in any alignment procedure is how to inject the generator signal into the receiver circuits. The problem is a dual one; the signal must be coupled into the receiver, yet it must



**Consumer Education.** Did you see "The Facts About the Cost of TV Repairs" in the June 28th edition of TV Guide? About 6,500,000 television set owners in the U.S. did! It warrants mention because it's just the sort of thing that helps clear up some of the mysteries the general public has about the TV repair business. Howard W. Sams & Co. has obtained permission from TV Guide to make reprints available for your benefit. If you'd like to distribute them to your customers, order from Dept. C.S., 2201 E. 46th St., Indianapolis 5, Ind., enclosing 25c per hundred to cover cost of handling and mailing. A sample of the reprint will be sent on request.

### \$ & ¢

**Right Dress.** "Servicemen should dress for a service call as though they were taking the lady customer out for dinner and a Broadway premiere. Suppose he does get a little dirty, poking around in the TV. She will think of Sir Walter Raleigh flinging his beautiful cape into the mud to provide a clean place for the feet of a queen. And who would quibble over a few dollars with so gallant a man?"

This tongue-in-cheek commentary from TEAM NEWS (published by The Electronic Association of Missouri) contains more than just a grain of truth. Neatness in dress is definitely appreciated by the customer, and making a good impression is worth the small effort it takes.

Whether you dress in sports clothes or in a uniform depends on the impression you want to create in the mind of the customer. A uniform suggests a businesslike, efficient organization. This is exactly the feeling you want to get across

in the case of a large shop, where the customer-technician relationship is fairly impersonal. Here, you actually try to create a "personality" for the shop itself, distinct from that of the technician. The uniform becomes a shop trademark, and the customer thinks, "Here comes A-1 TV" instead of "Here comes Charlie Brown."

The only trouble with a uniform is that some customers will associate it with gas-station attendants or route salesman, and mentally classify the technician as such. This is a handicap if you are trying to impress the customer with the fact that you are rendering professional services. Technicians in small shops, who depend most heavily on personal contact with customers, would especially benefit from upgrading their appearance by dressing up a little. Slacks and a well-pressed sport shirt are sufficient in many communities, but a dress shirt and tie may be preferred in some locations.

At any rate, dirty work pants and a wrinkled shirt will do nothing toward improving your standing with the customer — or toward raising your own impression of yourself.

### \$ & ¢

**Dogged Perseverance.** When a technician is hot on the trail of a defect in a "dog" TV set, time flies by unnoticed. It's easy to work through an entire half day or longer without realizing how much time has been spent on the job.

When you finally lay down your test probes and pick up a pencil to figure out the bill, how much compensation do you expect for putting in those long hours? You could charge your regular hourly rate for labor, but this could result in a bill of \$25 or more for replacing a few

small parts. At this point, it becomes hard to convince even your most understanding customers that you have earned your fee.

On the other hand, you could go to the opposite extreme and charge a flat rate for every type of repair — no matter how long it took you to complete it. The danger in this practice is the temptation to tell yourself, "I was too dumb to cure this trouble sooner, so it wouldn't be fair to charge for all the time I spent poking around in the set."

Charging some of your time to experience is all right, up to a point, but a technician is entitled to somewhat more than the average labor charge when he has to spend an unusual amount of time on a difficult bench job. After all, he must call on all his reserves of skill and patience to cure a trouble that is hard to find. To keep charges on long jobs from going sky-high, however, it may be necessary to average out the fees for different calls to some extent. That is, you should be sure that you allow enough profit on "easy" jobs to make up for losses on the inevitable "dogs."

### \$ & ¢

**Test Your Tact—VI.** When you run into an unexpected problem on a house call, it's only natural to stop for a moment and collect your thoughts before taking further action. If you are one of the many people who automatically reach for a cigarette at such a time, do you stop to find out whether the customer is willing to let you smoke in his home?

Before you ask, you would be wise to scout around the room for ashtrays. If they are lacking, you're probably better off not even to mention the subject of smoking. Fight off that "nicotine fit" until you can get out of the house! On the other hand, used ashtrays are a signal that the customer will probably OK your request for a few puffs.

By the way, how good are you at keeping ashes under control? It's a good trick even when your mind is not preoccupied with some technical problem. Unless your nicotine habit is absolutely uncontrollable or the customer offers you a cigarette, your best bet is to wait until you get outside for that satisfying puff.

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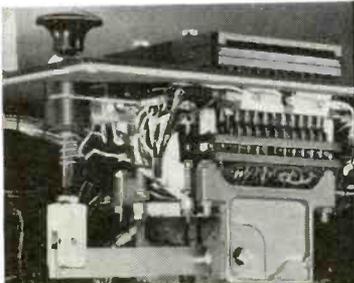
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# the troubleshooter

## answers your service problems

### Overloading

Since the flyback transformer on a Sparton Chassis 22V214 was replaced with a Merit HVO-9 and MWC-1, the set has had a tendency to overload on strong signals. The "no-signal" voltage at the grid of the AGC keying tube is only 90 volts, and the pulses at the plate of this tube have a peak-to-peak value of only slightly more than 400 volts instead of the 500 volts shown on the schematic. Is there some modification necessary in this AGC circuit because of the change in the flyback?

W. H. Guss

Roswell, N. Mex.

All voltages in the flyback circuit may be somewhat lower than they were before you replaced the transformer. As long as you still have sufficient width and brightness, however, you shouldn't need to be concerned with any alterations of the flyback circuit. The 400-volt pulses at the plate of the AGC tube are ample for normal operation.

On the other hand, an AGC grid voltage of only 90 volts is definitely too low to permit normal conduction of the keying tube. In this set, boost partly determines the voltage on the grid by means of the connection through R54; therefore, a decrease in boost voltage would cause a corresponding increase in bias on the AGC tube. Unless boost is so low as to hinder the operation of the sweep oscillators, though, I'd say it would be best to let it seek its own level. You can probably raise the AGC grid voltage by reducing the value of R54.

Low grid voltage on the AGC tube could also be caused by insufficient voltage in the plate circuit of the video output tube. (The latter condition in itself would cause overloading.) At least 120 volts should be present at the junction of R46 and R47 in the video circuit.

### Intermittent High Voltage

I've had five different calls to restore high voltage on a Crosley Chassis 473, and I restored operation in a different manner each time. The last two times, substitution of the damper tube did the trick; on another occasion, the horizontal oscillator tube had to be changed. One time I restored high voltage simply by bringing the set to the shop. How can I prevent this symptom from coming back?

GEORGE PATTERSON

Irwin, Pa.

This appears to be one of those ultra-exasperating intermittent conditions where the trouble won't stay around long enough to let you locate the source. A visual check of the horizontal and high-voltage circuits would be advisable, since the trouble could be something as simple as an intermittently open anode lead. I suggest a thorough check of the damper tube socket pins and connections.

If physical checks and tube substitutions don't result in a permanent cure, wait until the set acts up again; then check some of the waveforms in the circuit. The drive signal at the output-tube grid should be viewed first. Distortion of this waveform indicates trouble in the discharge-tube, oscillator, or AFC stages, and normal appearance indicates trouble in the output, damper, or high-voltage stages. Keep narrowing down your search until the trouble is found. Don't let the

set go back to the customer without trying your best to find a lasting remedy; nothing will hurt your business worse than repeated calls for the same complaint.

### Horizontal Weave

A Packard-Bell Model 2710 has an intermittent horizontal wiggle. When the set is first turned on, a bend appears in the picture and weaves slowly from top to bottom. The raster is not distorted, and I can remove the bend by reducing the contrast setting to a low value. The picture settles down to normal after five or ten minutes of operation, but the weaving reappears once or twice each evening.

R. J. GRANT

Culver City, Calif.

This trouble sounds very much like 60-cps hum interference in the video signal. Weaving occurs in such cases because the interfering signal is out of phase with the vertical sweep signal. Since the set in question has a full-wave power supply that produces 120-cps ripple, the most likely source of the spurious 60-cps signal is the heater circuit. Leakage from heater to cathode of some tube is frequently to blame.

You mention that the contrast control setting affects the bends. This could be a good, time-saving clue to the location of the defect. Your set has two stages of video amplification, with the sync and AGC take-off points in the first stage and the contrast control at the cathode of the second stage. This means that the contrast control should have no effect upon the operation of the sync or AGC sections of the receiver, and we can deduce that the trouble is probably not in either of those sections.

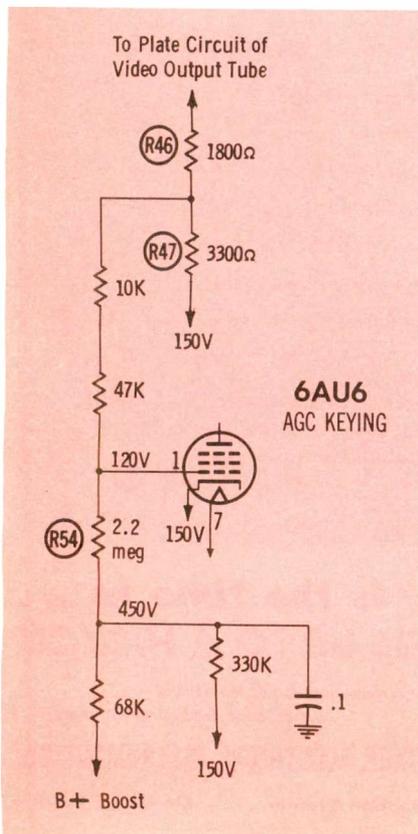
The most obvious suspect is the video output stage or the DC restorer that follows it. Try substituting (not just checking) the tubes in these circuits. If that does not cure the trouble, do some signal tracing with a scope. A humped appearance of the waveform (at a sweep frequency of 30 cps) corresponds to a bend in the picture.

### Horizontal Jitter

A Packard-Bell Model 21VT1 recently gave me quite a battle. The picture would vibrate horizontally at a fairly rapid rate. Vertical interlace also seemed to be affected at times. I did everything I could think of, and followed the suggestions of several others, but got no results. Then, for no special purpose, I checked a 10-mfd capacitor in the vertical output stage with my capacitor checker. (This component is connected to the bottom of the vertical output transformer and also through 12K ohms to B+ boost.) It showed a power factor of 20, so I slipped a new one in—and cured the trouble. Now I'm wondering why.

WADE L. NELSON

Highgrove, Calif.



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This electrolytic capacitor serves mainly to decouple the vertical output stage from the B+ boost line. The original unit was probably just defective enough to allow a spurious signal to appear across it at a 60-cps rate. Such a signal, when fed back to the horizontal output circuit by way of the boost line, would probably account for the jitter you observed. Congratulations on making full use of your capacitor checker. A thorough test appears to have paid off in this case.

## Bad Horizontal Drive

A DuMont Chassis RA-340 had sound but no picture, and the plate of the 6CD6 horizontal output tube was red hot. When I advanced the horizontal drive control, the picture came in. Everything returned to normal except that a drive bar appeared near the center of the screen. The customer noticed this and called me back to remove it; however, I could not do so without bringing back the original trouble. I wonder if the control might be bad, or if there might be some other trouble with the circuit?

MICHAEL J. SHIVEY, JR.

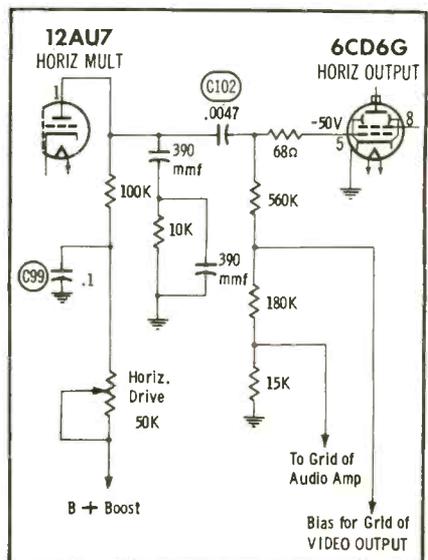
Cleveland, Ohio

Since you are unable to get a strong enough drive signal without distortion, something is apparently wrong with the second section of the horizontal multivibrator or the coupling circuit between the multivibrator and output stages. There might possibly be open spots in the drive control, but I would sooner suspect trouble in some other component. Here are a couple of other possible defects:

1. The second section of the multivibrator tube could be too weak to supply a normal drive signal.

2. C99 could be leaky, thereby reducing the plate voltage at pin 1 of the multivibrator tube and thus reducing output signal amplitude.

If you can find no defects in the multivibrator plate circuit, check the grid and screen circuits of the output tube for possible troubles.



## Police Radio Interference

The local State Highway Patrol radio station interferes with reception on a Motorola Model 21K53MA. Each time a call is sent out, the TV sound is distorted and the picture tears. Both local stations (Channels 7 and 9) are affected by this interference. Frequency of the Patrol transmitter is 42.62 mc.

I. C. WYATT

New Bern, N. C.

Since the interference falls right in the video IF band, it'll be fairly tough to get rid of without seriously reducing picture fidelity. About all you can do is to trap out as much of the interfering signal as possible. You might first install a high-pass filter between the antenna and the terminals on the tuner. If the interference is still objectionable, you might try installing an extra 41.25-mc sound trap (or another unit with very high Q) and retune it to exactly 42.62 mc. A series resonant circuit from the grid of the first IF to ground might do the trick, but if it doesn't give good results in this location, you might try moving it to one of the other IF grid circuits. More than one trap might be needed.

By keeping the Q of the trap extremely high, you will minimize the loss of video information. Fortunately, the portion of the IF band in the vicinity of 42.62 mc corresponds to the higher video frequencies; this means that adding the extra trap should not cause picture degradation other than the loss of a certain amount of fine detail.

## Wobbly Dots

I have an RCA WR-36A white dot generator and an Admiral C322C2 color TV. When I try to perform convergence with any of the four patterns provided by the generator, the pattern wobbles in time with vertical retrace lines that run up or down on the screen. The lines can be temporarily stopped by adjusting the hold control on the TV or the vertical bar control on the generator, but they cannot be held stationary. They are very bothersome in most convergence work because their speed and direction of motion is constantly changing. How can they be eliminated or controlled?

L. U. KING, JR.

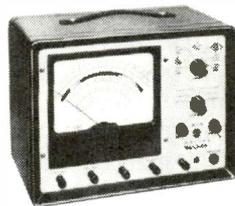
Corning, Ark.

This generator provides two methods of synchronizing the pattern with the TV receiver's sweep oscillators. In the "Line" position of the sync switch, synchronization depends on power line frequency. The vertical hold control of the receiver must usually be adjusted in order to lock in the pattern. If stable sync cannot be achieved by this method, try placing the sync switch in the "External" position and feeding sync pulses from the TV set to the generator. The latter method, which is more positive-acting, is easy to use by following the explicit directions given in the instruction manual.



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AC (peak-to-peak)	0.2 to 4200V	0.2 to 4200V	—
Resistance	0 to 1000 meg.	0 to 1000 meg.	0 to 1000 meg.
Current	10 µa to 15 amp.	—	—
Accuracy (full scale):			
Voltage:			
DC (+)	±3%	±3%	±3%
DC (-)	±3%	±3%	±5%
AC	±3%	±3%	±5%
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# QUICKER SERVICING

By Calvin C. Young, Jr.

## New-Type CRT

A recently-introduced picture tube design that features low-voltage operation of grid number two (low Eg2) can be a source of trouble to the serviceman who encounters one without warning. As of June, 1958, there were seventeen low-Eg2 tubes registered with EIA, including five 14" types, six 17" types, four 21" types, and two 24" types. All of these units feature 90° deflection, 12L basing configuration, and low-voltage electrostatic focusing. Six of them are designed to operate with a potential difference of 110 volts between G1 and G2—the others with only 50 volts difference. Referring to EIA basing diagram 12L in Fig. 1, note that G1 and G2 as mentioned here are the control grid (pin 2) and accelerator grid (pin 10), respectively.

After mentioning some of the major reasons for the development of this tube, we will examine in detail the principles that make this type of operation possible. Recently, there has been some discussion of picture tubes which can be driven directly by the video detector circuit, completely eliminating the need for a video output stage. That there has

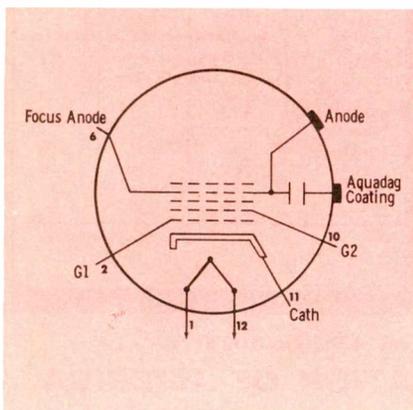


Fig. 1. EIA picture tube basing configuration 12L applies to low-Eg2 types.

been progress in this direction is borne out by the introduction of the low-Eg2 tube, because it requires less video drive to achieve the same contrast ratio produced by earlier tubes operating with 300 volts or more on G2. This means that it is now possible to obtain a higher contrast with the same video drive, or the same contrast ratio with less drive, a very important consideration for portable TV receivers having contrast ratios which leave something to be desired. Another important feature is that G2 can be powered directly from the low voltage supply, even in a set that employs a half-wave power supply having a maximum DC output of only 125 to 135 volts.

With these advantages in mind, let's see how the low-Eg2 tube operates. In theory, it can be compared to a pentode receiving tube, where plate current is relatively independent of plate voltage as long as it exceeds the screen voltage. Plate current is a direct function of screen and bias voltages, the latter having the greater effect.

In a picture tube, G2 can be compared to the screen grid of a pentode and the ultor anode to the plate. The additional focusing and acceler-

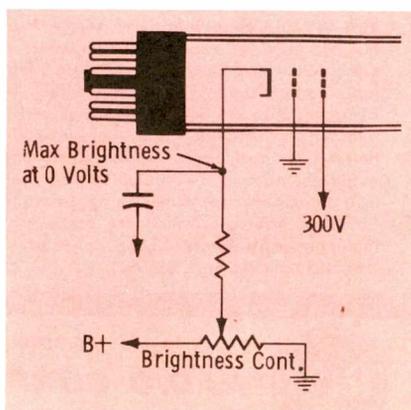


Fig. 2. Basic CRT circuit showing bias arrangement for maximum brightness.

ating anodes used in the gun structure do not affect this comparison since their function is to concentrate the electrons into a narrow, rapidly-moving beam.

Having established the mode of operation, let's proceed to see how the low-Eg2 tube applies these principles. Consult the schematic of a basic picture tube gun structure in Fig. 2 and note that for a standard CRT, the first anode (G2) operates at about 300 VDC, the control grid at ground potential (OV), and the cathode receives the video driving signal. In addition, the brightness control is connected into the cathode circuit to provide variable bias. The video driving signal (for cathode drive) is positive-going, and maximum brightness is obtained when the cathode voltage equals the control grid voltage. In this configuration, the positive-going video signal increases instantaneous bias and causes a corresponding change in the instantaneous brightness level. Because the setting of the brightness control determines the fixed bias on the picture tube, maximum illumination is a function of the brightness-control setting when all other voltages remain unchanged. Maximum contrast, however, is a function of the amplitude of the video drive signal, increasing in more or less direct proportion until saturation and cutoff levels are reached. To summarize, we can say that maximum usable brightness and contrast levels are obtained when the DC bias is such that the video blanking pedestal causes raster cutoff and maximum white produces a condition of zero bias.

Looking at a tube manual, we find that raster cutoff for a 14NP4 occurs between 33 and 77 volts, while cutoff for the equivalent low-Eg2 tube (a 14AEP4) is obtained somewhere between 37 and 55 volts. If these tubes are operated at their recommended G2 values of 300 and 110 volts respectively, and are cathode driven, the video drive requirements for raster cutoff are 48 and 42 volts, respectively. From this comparison alone, it can be seen that the low-Eg2 tube type requires less video drive to produce the same contrast ratio.

To understand how this increase in efficiency is possible, let's examine the basic circuits of the two tubes

**Sylvania comparisons point out—**

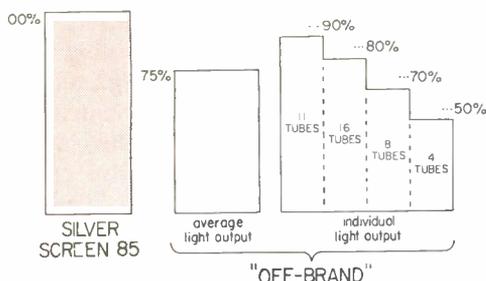
# The big difference in Picture Tubes!

Here's the inside story on why local "off-brands" don't measure up to Silver Screen 85® standards

If you're like most dealers, you know off-brand tubes don't have the same quality standards as first-line tubes. To help you see how big the difference is, Sylvania purchased a nationwide sample of sixty 21YP4A's made by 19 different local tube makers. These tubes were put through the same production tests that all Sylvania tubes must pass.

**Not a single local off-brand passed all 54 mechanical and electrical tests!** Many of these were minor defects making little or no difference in whether or not the tube "lit up." But look how loose manufacturing controls can affect the important features of light output, focus, and life!

## LIGHT OUTPUT



So far, 39 off-brand tubes have been compared with the *minimum* light output of Silver Screen 85. Five additional tubes couldn't even be tested. Eleven tubes were less than 90% as bright as the minimum for Silver Screen 85; 16 were less than 80%; 8 were less than 70%; and 4 were *less than 50%* as bright. Since most Silver Screen 85 tubes average as much as 125% of minimum standards, the difference becomes even greater. Small wonder that Silver Screen 85 is the easy way to more satisfied customers.

## FOCUS

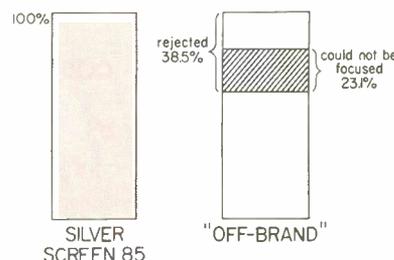
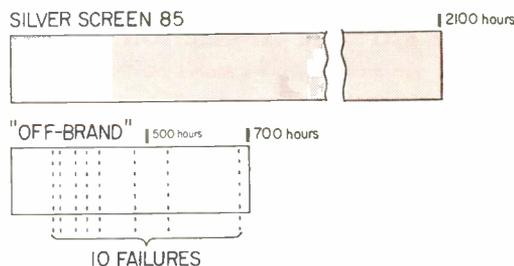


Chart 2 shows how these same 39 tubes stacked up to registered limits on focus voltage. 38.5% were rejected under these limits. Over half of all those rejected could not be focused in a TV receiver. Small wonder then that "Silver Screen 85" pictures are sharper, brighter, clearer.

## LIFE TEST



Nineteen off-brand tubes were placed on Sylvania's standard 2000-hour life test. Chart 3 tells you how fast these tubes developed slow-heating cathodes. Over half, or ten units, failed to go beyond the 700-hour mark. Small wonder then that Silver Screen 85 gives you less troublesome callbacks.

Of all the off-brand tubes tested, Sylvania engineers estimate that 43% probably would not have operated properly in a TV set. Why gamble your reputation, customer satisfaction, and success. It's just *good business* to sell up to "first line" picture tubes; Silver Screen 85 picture tubes.



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as shown in Fig. 3. In both A and B you will notice that the video drive signal is 42 volts. (This value was used for comparison purposes since the 48-volt signal required by the 14NP4 would be excessive for the 14AEP4 tube.) If we assume that a bias of 10 volts is required to achieve optimum video drive, the potential on the cathode will increase by 32 volts with the presence of the blanking pulse. For the 14NP4, this means that the cathode will be, at that instant, 32 volts closer to G2, or that the difference between cathode and G2 will be reduced from 290 volts to 258 volts.

Now if we consider the other case, where the grid has a potential of 25 volts and the cathode 35 volts (again providing a 10-volt bias), we have a difference of 100 volts between the cathode and G2. With 42 volts of video drive, application of the blanking pulse will reduce the voltage between cathode and G2 to 68 volts. It can be further seen that the change from 100 to 68 volts is greater percentage-wise than the change from 290 to 258, even though the change was 32 volts in both cases. Viewed in this light, the increased efficiency (reduced video drive required for same contrast ratio) is easily understood. Low-Eg2 operation is possible because of the improved electron gun assembly. We say improved rather than new because the electron gun used in low-Eg2 tubes has the same general configuration as the standard picture tube.

**What This Means to Technician**

Aside from the improved performance features of low-Eg2 tubes, there are at least two other associated factors that will be of interest and concern to you. First and foremost is testing of the new tube. Naturally, you wouldn't expect to test it with 300 volts on G2. (The absolute maximum voltage rating between G2 and G1 on the 14AEP4 is 220 volts and is as low as 68 volts on some tubes in this same general class.) However, as one might suspect, with certain modifications to the test setup procedure, existing CRT testers can be employed. At least one manufacturer of test equipment, when contacted about this, revealed that revised instructions on

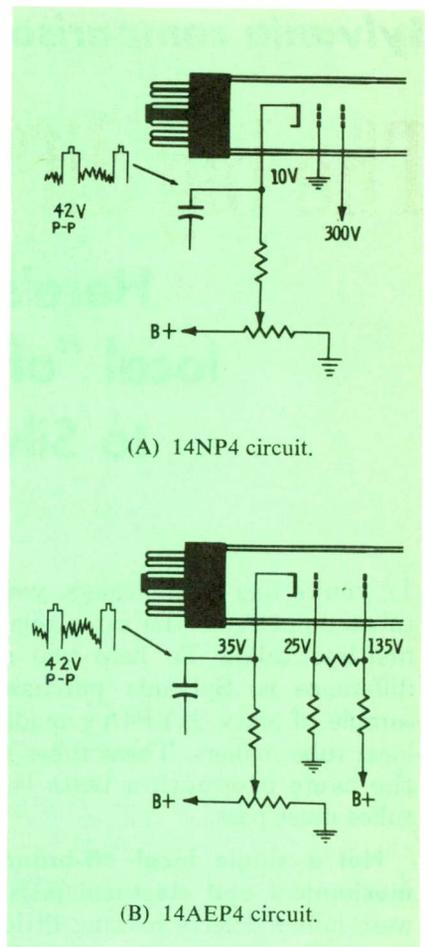


Fig. 3. Comparison of bias arrangements, standard and low-Eg2 circuits.

testing low-Eg2 tubes were being sent to owners of their brand CRT tester. However, the instructions carried a warning to the effect that testing low-Eg2 tubes in any but the prescribed manner could ruin the tube. With this in mind, it is suggested that you contact the manufacturer of your test instrument for instructions on testing these new type tubes. Further, do not test these tubes until you have received instructions, studied them, and are absolutely sure you understand just what to do.

Also of interest is the fact that the new tube will replace its standard equivalent with only minor circuit changes. If a receiver circuit includes cathode drive and a cathode brightness control, the low-Eg2 tube can be installed by simply changing the voltage on G2 to the correct level. Fig. 4 is a basic circuit diagram of a low-EG2 tube. Note that the control grid is operated at a positive potential derived from the same resistive voltage-divider network which supplies G2. This is done be-

cause the potential difference between G1 and G2 must remain fixed. Notice also that a vertical-retrace blanking signal is coupled into the control grid circuit. The pulse must be negative-going and have an amplitude of about 80V p-p.

For the signal condition shown, 50V p-p, the picture tube is being operated with 10 volts of bias between the cathode and the control grid. This is necessary to prevent excessive contrast, and indicates that a 40V p-p signal would, in all likelihood, produce adequate contrast.

It should also be apparent from Fig. 4 that the values of R40, R41, and R42 are such that they carry a current of about 1 ma (270 volts across 263,000 ohms produces .975 ma). This means that if you make a similar resistor choice, the voltage drop can be determined without effort. For instance, if your power supply produces 135 volts and you need 50 volts between G1 and G2, use a total of 135,000 ohms for the three resistors and values of 25K for R40, 50K for R41, and 60K for R42. You may want to select the nearest EIA standard values — say 22K, 56K, and 56K, respectively. The same principle can be applied to any supply voltage; however, do not supply this network from B+ boost.

### Magnetize or Demagnetize Tool Instantly

The Perma-Power "Magneformer" (Fig. 5) should be a boon to the service shop. Tools become magnetized when removing or installing permanent-magnet focus assemblies

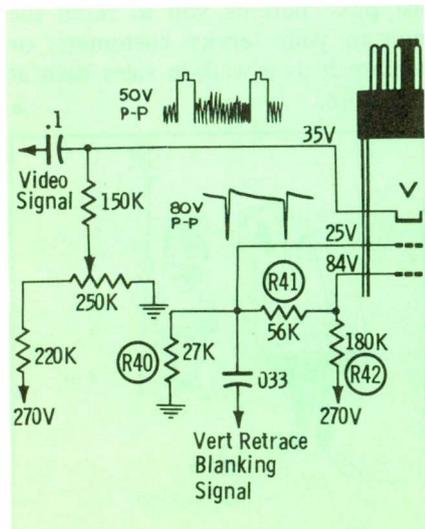


Fig. 4. Practical voltage-divider circuit for low-Eg2 operation of picture tube.



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Fig. 5. "Magneformer" can be used to magnetize or demagnetize hand tools.

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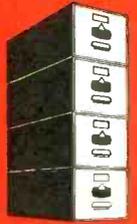
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# servicing the summer AIR-CONDITIONER

Maintenance Data for Radio-TV Servicemen by Joseph Derman

Summer air improvement, both in the home and in public places, has almost become a necessity. The TV set owner may already have an air-conditioner, or if not, is very likely considering the purchase of one. Shopping centers, movies, offices, and other such areas are employing air-conditioning equipment at an ever-increasing rate. Since the popularity of air-conditioners during the warm season naturally complements other types of service during the balance of the year, the electronic service technician would do well to weigh the possibilities of servicing air-conditioning equipment. This article will familiarize the reader with air-conditioners based on the principles of mechanical refrigeration, and will then examine some major troubleshooting and testing techniques.

Mechanical refrigeration is basically the same process, whether applied to the air-conditioner, the household refrigerator, the freezer, or any other kind of cooling device. It can be defined as the controlled evaporation and condensation of a suitable liquid (refrigerant) in a properly designed system.

We are all familiar with the cooling effect developed by the evaporation of a drop of alcohol from the skin. In normally encountered temperatures and pressures, alcohol will evaporate and become a gas, thereby causing an amount of heat energy to be expended. Since, in this case, the skin provides the required energy, it becomes cool.

The air-conditioning unit (Fig. 1) accomplishes cooling in much the same way. A measured amount of a special liquid is permitted to evaporate,

thus extracting heat from the space to be conditioned. The component in which this actually takes place—the evaporator—resembles the radiator of an automobile. It is composed of a system of copper tubing equipped with fins to provide a large area of contact between the evaporating liquid in the tubing and the surrounding air. A blower brings room air into the air-conditioner to aid in the evaporation process and thus cool an increased quantity of air.

For obvious reasons, it is desirable to reuse the refrigerant liquid. The air-conditioner does this as shown in Fig. 2. The evaporated liquid (gas) from the evaporator is drawn into the compressor where it is compressed in order to raise its temperature. In this condition of elevated temperature and high pres-

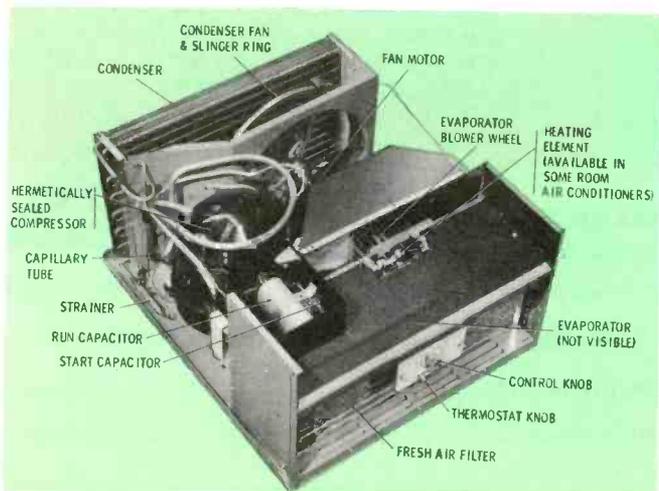


Fig. 1. Component layout in typical window air-conditioner.

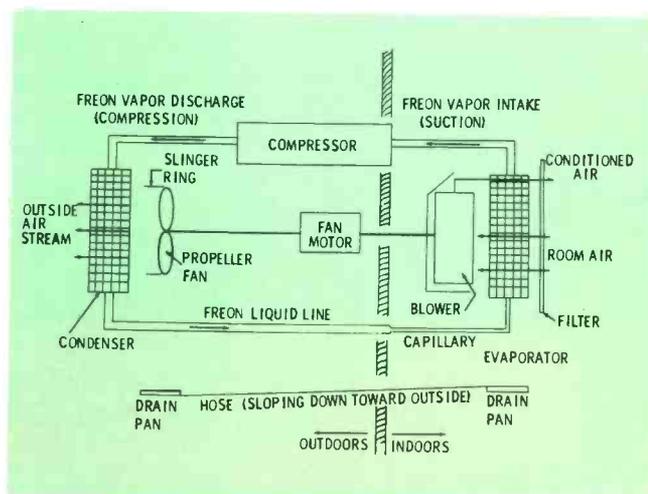
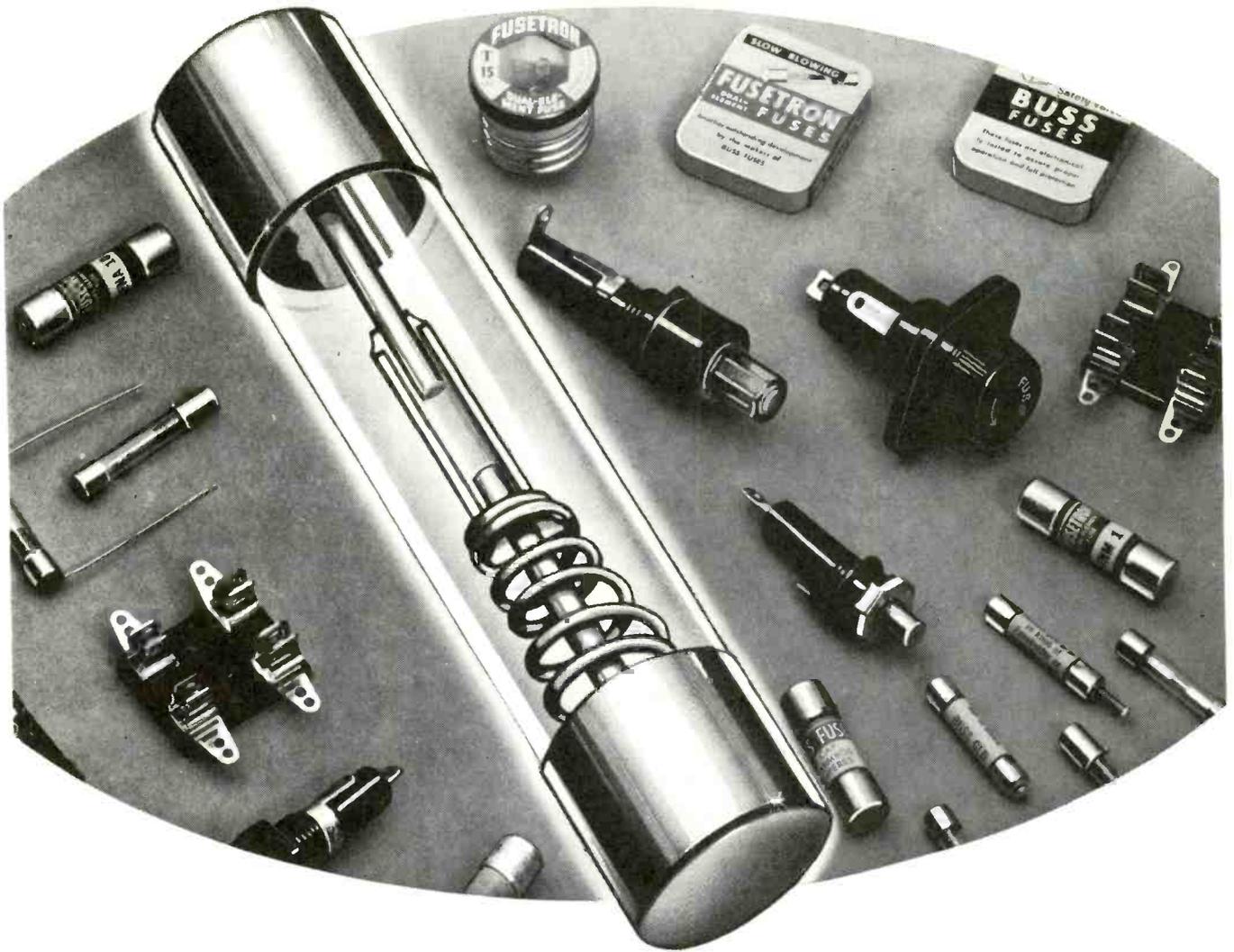


Fig. 2. Block diagram of air streams and refrigerant system.



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sure, the gas is forced into the condenser which is located outside the conditioned space. The construction of the condenser is similar to that of the evaporator and facilitates another heat transfer. The heat of the compressed gas is liberated to the outdoor air, causing the gas to liquefy. The liquid is now ready to start another cycle of evaporation, compression and condensation. To control the rate of entry of the refrigerant into the evaporator, most manufacturers of window units use a capillary tube comprised of a few feet of small-bore copper tubing, in series with the refrigerant circuit.

The air streams (both inside and outside the room) contribute to the efficiency of the refrigerating process. The blower pulls room air through the filter as well as through the evaporator and then directs it back into the room. Since the filter removes dust, pollen, etc., the air is cleaned in addition to being cooled. Dehumidification, or the removal of moisture from the air, also occurs at the evaporator because air supports less moisture as it cools. The moisture that condenses out of the evaporator fins and tubing is collected and piped from a pan under the evaporator to another pan beneath the condenser fan.

Removal of moisture from the room being conditioned is important because the body cannot dispose of its perspiration readily in an atmosphere of high humidity. Evaporation of perspiration (and consequent body cooling) increases as moisture in the atmosphere decreases.

The function of the propeller fan is to bring outdoor air through the condenser in order to cool it and thereby aid in the condensation of the refrigerant. To increase the effectiveness of the condensing action, most manufacturers make use of the condensate (water) that is collected at the evaporator. This is usually done by equipping the propeller fan with a lip at the tips of the fan blades. As the fan rotates, this "slinger ring" scoops up the condensate and sprays it against the condenser to aid in cooling the fins and tubing.

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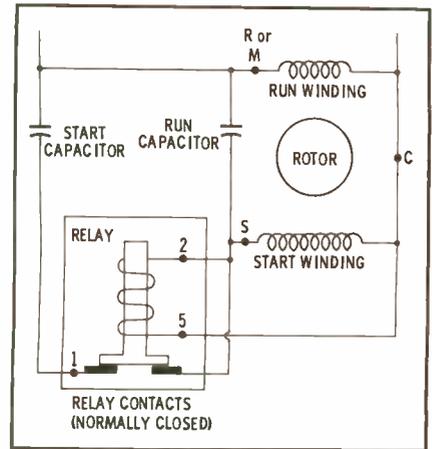
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**Fig. 4. Start-capacitor contacts will open when motor comes up to rated speed.**

shading winding becomes ineffective and the main winding alone is sufficient to maintain rotation. A single-speed fan motor will have one stator winding; two-speed fans require two stator windings.

Considerably greater motor torque is required to operate the compressor than can be developed by the shaded-pole type of motor; consequently, an induction motor of the capacitor-start type is generally found in a motor-compressor system (see Fig. 3). The necessary torque for rotor starting is established by the phase difference between the current in the main winding and that in the start circuit (start winding and start capacitor).

An ohmmeter can easily be used to distinguish one winding from the other. The run winding, which is designed for continuous operation, consists of a relatively-small number of turns of heavy gauge wire; but the start winding, operated only for the short period during starting time, is made up of a larger number of turns of thin gauge wire. The resistance of the start winding, therefore, is generally greater than that of the run winding.

The stator windings (that is, the run and start windings) are an integral part of the sealed motor-compressor unit. Only the terminals — "C" (common) "S" (start), and "R" or "M" (run or main) — are accessible on the outside of the compressor. Both start and run capacitors are external to the compressor. The use of the start capacitor is an improvement over earlier types of induction motors because it provides for greater starting torque. The running capacitor is a recent develop-

ment that helps to improve the power factor of the air-conditioner load.

It is the usual practice for the start circuit to be disconnected as soon as the motor reaches about 85% of rated speed. Disconnect devices used in air-conditioning, called starting relays, are available in either current or voltage types.

In the current relay (Fig. 3), the initial surge of current to the main winding closes the normally-open contacts and inserts the starting circuit. As the speed of the motor increases from stalled (stationary) position to rated speed, current to the motor decreases. Then the relay contacts open and disconnect the start circuit. The run winding alone is sufficient for continued operation of the motor.

The voltage relay (Fig. 4) has contacts that are normally closed, which means that the start circuit is already connected before power is applied to the motor. As the rotor comes up to the normal speed, the relay is energized sufficiently to maintain the contacts in an open position, thus removing the start capacitor from the circuit.

Two control elements found in most air-conditioners are a thermostat and the thermal overload protection device. The thermostat is an automatic switch that turns the compressor on or off to maintain the temperature of the conditioned air within a preselected range. A gas-filled tube, suspended in the return stream of room air, actuates the make-and-break device that controls the operation of the compressor. The fan motor is independent of the thermostat.

The thermal overload is an automatic control designed to protect the compressor motor against excessive current. An important characteristic of induction-motor operation is that the current decreases from the maximum initial current surge to a much lower normal load current. If, for any reason, the motor does not turn over, the current drain will be excessive and the thermal overload will open the compressor circuit. In a short while, however, the overload unit will again permit the motor to go on. Motor operation will then continue normally if the cause of improper motor operation has been removed.

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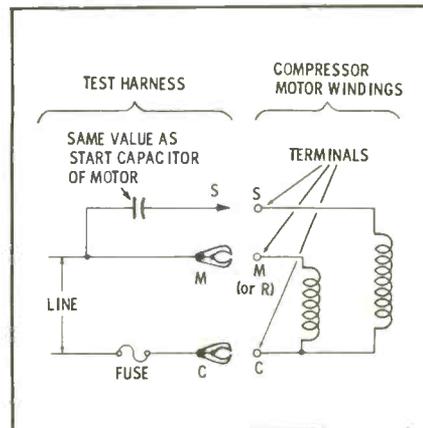


Fig. 5. Compressor-motor test harness includes substitute for start capacitor.

### Troubleshooting Problems

When an air-conditioner service problem is being analyzed, the trouble should first be localized to the electrical, refrigeration, or air-circulation system. Then it can be further analyzed in the following manner:

#### Electrical Section

*Symptom:* Unit is inoperative because fuse has blown, or fuses blow repeatedly.

*Analysis #1:* Initial current surge of normally - operating induction motor is great enough to open an ordinary fuse.

*Solution:* Use a fuse of the thermal-delay type. This type of fuse is designed to pass the initial current surge for induction-motor starting and yet afford protection against a sustained short or current overload.

*Analysis #2:* Low line voltage at the unit prevents the compressor from starting. Under this condition, excessive current is drawn and even the thermal-delay type of fuse will blow.

*Solution:* The voltage applied to an induction motor should be within 10% of the rated value. Low line voltage may exist at the unit because of heavy power demand in the community, increased local demand on the same branch circuit for other appliances, or high IR drop between source of power and unit due to inadequate wire size or too great a distance. It is advisable to check the adequacy of available power by measuring the voltage across a 1000-watt resistive load at the outlet to be used by the air-conditioner. If other appliances are overloading the air-conditioner power circuit, they should be removed. Only #12

or larger wire should be used for extension cords when these are necessary.

**Analysis #3:** The compressor motor does not turn over and therefore draws high current.

**Solution:** Compressor failure may have many causes such as a defective winding (open or short), a short to chassis, or a defective start capacitor. Use an ohmmeter to check for winding continuity. (The run winding may have a resistance of approximately 3 ohms, while the start winding may have a resistance of approximately 15 ohms.) Test for a short between the compressor terminals (C, S, and M or R) and chassis. There must be no indication of continuity between the terminals and the compressor shell. Check the start capacitor with an ohmmeter. Use an exact replacement, if the capacitor is suspected of being defective.

Occasionally, one may encounter a "stuck" compressor which will not operate because of internal mechanical reasons. It had best be replaced.

Fig. 5 shows a test harness with which to check the compressor motor. To use it, disconnect other components and use a capacitor of correct value required to start the motor. With the "M" and "C" leads already connected, apply the start-capacitor lead for a few seconds only. The motor will start up if it is satisfactory. (Note: The above discussion emphasizes high current drain caused by compressor-motor failures because this motor requires as much as 10 times the rated current of the fan motor.)

**Symptom:** Compressor motor does not operate at all.

**Analysis:** Power may not be reaching the compressor, or the compressor may be defective.

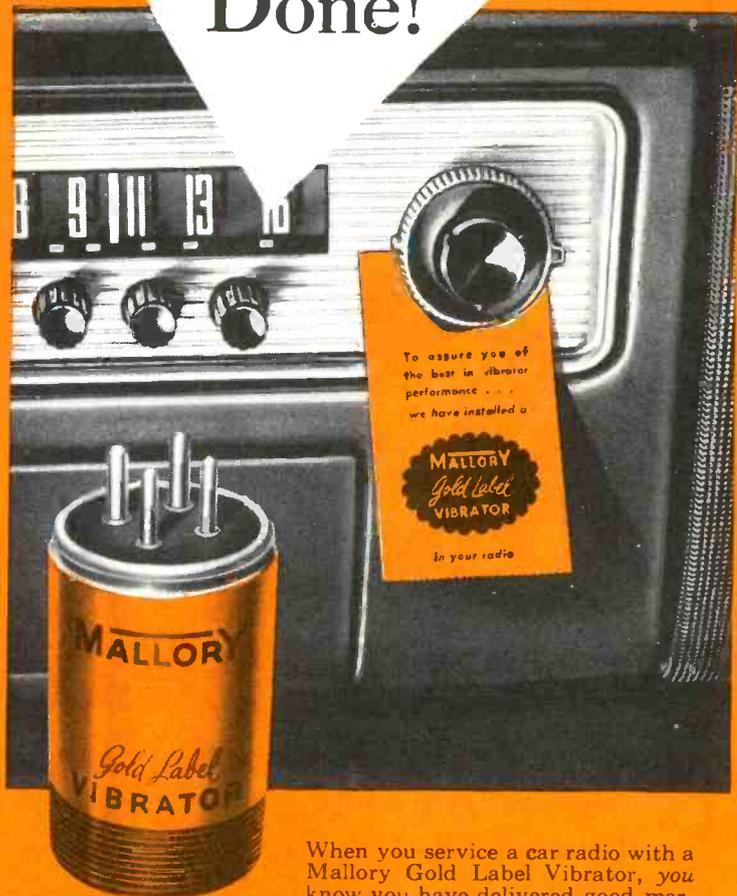
**Solution:** Check voltage at outlet, switch, cord, etc. Use the compressor test harness. There's a possibility that the room temperature is not high enough to close the thermostat contacts. This condition can be verified by advancing the thermostat control so that the unit will "cut in" at a lower temperature, or by shorting across the thermostat with a jumper wire.

The thermal overload also may be defective. Use an ohmmeter to check for continuity across contacts and through the heater element of the

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overload control. *Important:* Do not operate the unit without an overload protector. Check the relay coil and contacts; if you replace the relay, also replace the start capacitor.

*Symptom:* Compressor cycles (goes off and on repeatedly).

*Analysis:* Automatic thermal overload makes and breaks the circuit—it is doing its job.

*Solution:* A condition such as unusual heat of the sun on the outside of the air-conditioner may increase the current through the overload until that unit is actuated. Place an awning over the air-conditioner in this case, being careful not to block outside circulation around the condenser. A partial short in the electrical system or a marginal low-voltage condition will contribute to increased current drain and therefore to a condition of overload.

Only a brief mention can be made of the operation of the fan motor and its blades. Both the blower wheel and the condenser fan must rotate freely and at rated speeds for adequate air-conditioner performance. Does the motor require maintenance, i.e., oil, etc.? Do the fans slip on their shafts? Does anything keep them from rotating freely? A slow or stopped condenser fan may reduce refrigerant condensation enough to raise the head pressure in the condenser side so as to cause current overload. The compressor is then forced to work harder against increased pressure.

### Refrigerator System

*Symptom:* Evaporator coil freezes up.

*Analysis:* The reduction or loss of any considerable amount of Freon from the system results in evaporator pressures lower than the unit is designed for, causing excessive local cooling. The same condition of freeze-up can be caused by an obstructed air stream.

*Solution:* Examine the air stream (inside and outside) for dirt or any other obstruction. Replace or clean the filter. See whether the fan motor or fans are operating at reduced speeds. Check the line voltage. The unit will have to be pulled into the shop if a gas-leakage test or recharging is needed. Observe the manufacturer's warranty rules with respect to breaking into the sealed system.

## Air Stream

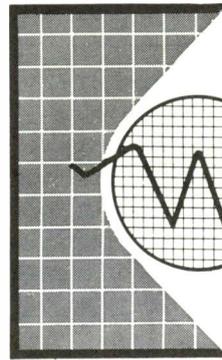
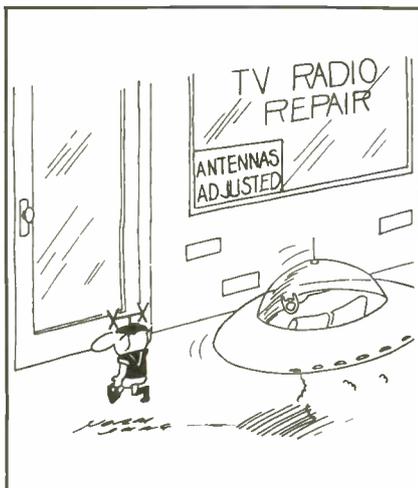
**Symptom:** Ineffective cooling, freeze-up of evaporator, high wattage consumption, etc. (Wattage consumption must be checked against manufacturer's specifications.)

**Analysis:** The air stream at the evaporator and condenser must be free and unobstructed. Any interference with air flow prevents evaporation of the correct amount of *Freon*. The temperature of the room air must be warm enough to furnish the required energy for *Freon* evaporation; on the other hand, an excessive load (or room air much too hot) will interfere with operation because of too much *Freon* evaporation.

The condenser air streams must also be free-flowing and not too hot. Unusually high temperature may trigger the overload, reduce condensation of the *Freon* gas and cause high head pressure.

**Solution:** The filter must be clean, and the blower wheel and condenser fan must rotate freely and at proper speed. Oil and clean the fan motor as necessary. The outside louvers must be clear of obstruction, and the condenser fan and slinger ring must be adjusted to pick up enough condensate to hurl against the condenser. Check to see that the unit is tilted so that condensate can collect under the condenser fan.

This brief analysis will serve to illustrate that troubleshooting problems and servicing techniques in the air-conditioner field are not greatly different from those in other electronic work. The television technician should feel right at home in air-conditioner servicing, once he becomes familiar with this type of work. ▲

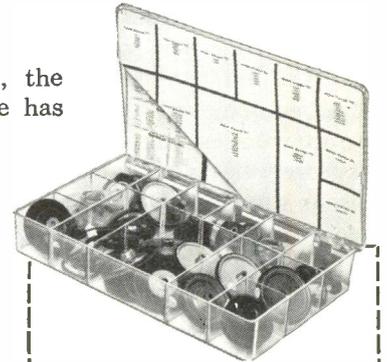


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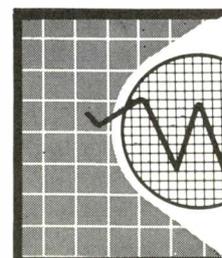
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## Horn-Type Loudspeaker Stereo Preamp Design

By Calvin C. Young, Jr.

In the two previous columns, we examined cone-type speakers in detail. The limited performance of the direct-radiating speaker at the low frequencies and its low efficiency over the entire range are two reasons why horn-type speakers are used in many commercial, PA, and hi-fi systems. In this article, we will examine the basic principles of horn-type speakers and see just how various aspects in their design affect operation.

### Basic Horn Styles

A horn-type loudspeaker consists of a driver assembly (sound generator) and a tube of varying sectional diameter (horn). The most common styles of horns (Fig. 1) are conical, exponential, hyperbolic, and parabolic. The flare of a horn may be either single or compound (Fig. 2), depending on the propagation or distribution pattern desired. Generally speaking, a horn which has a compound flare (equal in all directions) will have a fairly uniform distribution pattern, while a horn having only a simple flare will have a wider distribution pattern along the smaller axis of the mouth opening.

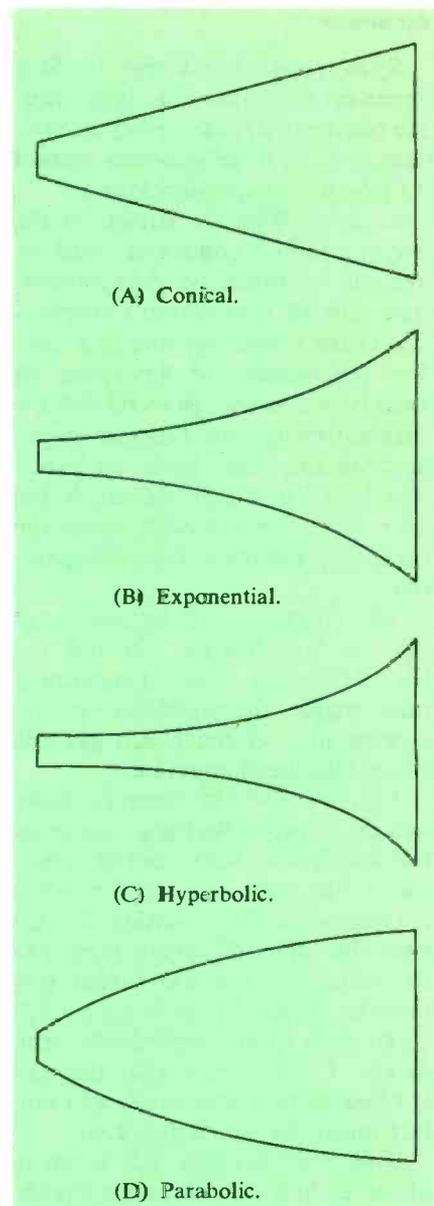
High- and low-frequency horns may use any of the styles shown in Fig. 1; however, a low-frequency horn will be large (in length and mouth diameter), while a high-frequency horn will be relatively small. When we speak of a high- or low-frequency horn, we are speaking of the cutoff characteristic. A large horn can produce the higher audio frequencies; but, unless it is assisted in this respect, considerable distortion can be developed in the upper range. Good low-frequency performance demands a large mouth area and a small rate of taper. A recent engineering textbook states that a taper constant of .1 produces a horn which doubles in area each 7.5" and has

a cutoff frequency 100 cps lower because of the taper. Further, cutoff frequency decreases as horn length increases (assuming the same rate of taper). With this factor in mind, it is easy to visualize that a low cutoff frequency is obtained only with a long, gently-tapering horn which has a large mouth area.

It seems logical to assume that a high-frequency horn would be short and have a small mouth area, or that best high-frequency response would be obtained by eliminating the horn entirely. This is true — but only up to a point. Because of efficiency considerations, the horn cannot be completely eliminated. A horn-type unit is more efficient than a cone-type because its diaphragm works into the resistance of the horn throat; the cone-type (in free air) works only against mechanical inertia. In other words, a small diaphragm in a horn-type speaker acts against the large column of air contained within the horn, while a cone works only against the air it contacts. Thus, for the sake of efficiency, it is necessary to have a horn of some size even for the higher-frequency units.

The extreme lengths required for low-frequency horns are obtained through the use of folded designs. In this way, effective horn lengths of 9' to 12' can be contained in a unit only 3' to 4' long — or even shorter, depending on the number of folds. Folding a horn causes some loss in performance due to the difference in path lengths along the inside and outside turns. Losses can be held to a minimum by confining the bends to the sections where the horn area is small; however, bends cannot be used in horns designed to produce frequencies above about 5,000 cps.

Horns may be circular or square with no appreciable change in performance. If a square horn is used, however, corner fillets must be em-



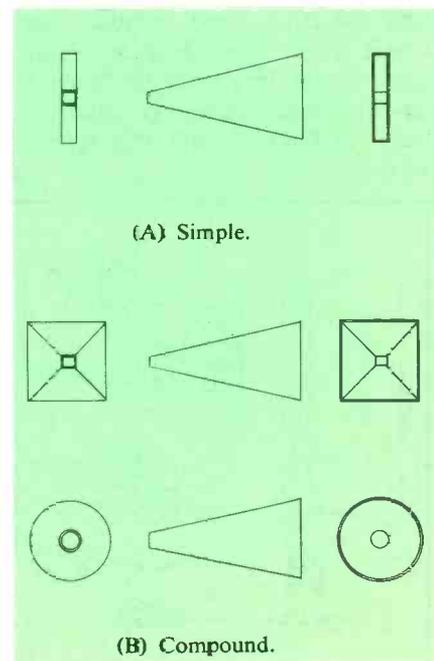
(A) Conical.

(B) Exponential.

(C) Hyperbolic.

(D) Parabolic.

Fig. 1. Horn flare styles.

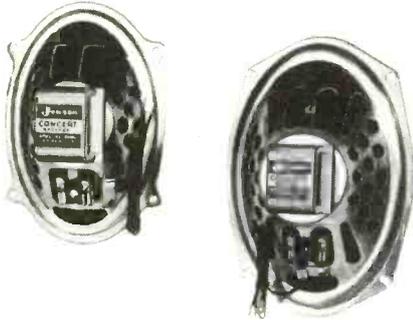


(A) Simple.

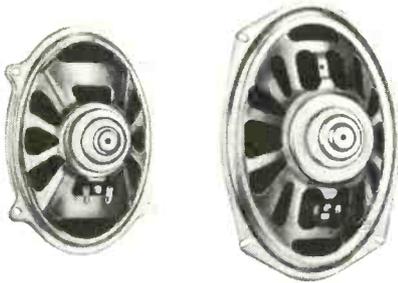
(B) Compound.

Fig. 2. Horn flares.

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5 x 7	P57-TX§	2.5‡	3.2	9.0	1	5 x 7¼	2½	4¼ x 6½	8.35
6 x 9	P69-U	1.73†	3.2	9.0	1	6¾ x 9½	3½	5¼ x 8½	7.80
6 x 9	P69-T	2.5‡	3.2	10.0	1	6¾ x 9½	3½	5¼ x 8½	8.65
6 x 9	P69-TX§	2.5‡	3.2	10.0	1	6¾ x 9½	3½	5¼ x 8½	8.95
VIKING SERIES									
5 x 7	57-J9	1.47	3.2	6.0	¾	5 x 7¼	2½	4¼ x 6¼	5.85
6 x 9	69-J9	1.47	3.2	7.5	¾	6¾ x 9½	2½	5¼ x 8½	6.50

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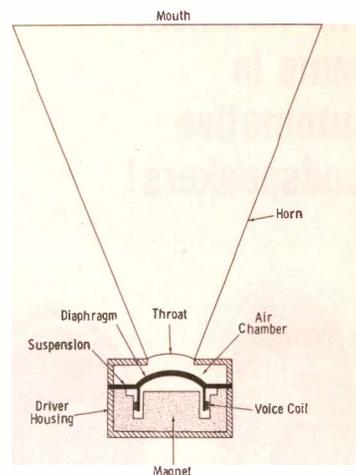
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ployed in the throat end to provide a smooth transition between the square section and the round diaphragm normally used in the driver unit. The relationship between the horn and the driver is shown in Fig. 3. Materials in common use for horns include wood, metal, and molded fiberglass. All of these materials must be used in sufficient quantity, and be adequately braced, to prevent the horn walls from vibrating and causing distortion.

In the early days of audio, before too much was known about horn action, conical horns were often employed. However, an exponential horn having the same length and effective mouth area as a conical design has been found to be superior in performance. This is due, at least in part, to the difference in throat impedance characteristics. The throat impedance of a conical horn rises from a low value of about .05 acoustical ohms at 200 cps to about .75 at 7000 cps; in an exponential horn, impedance rises from 0 to .95 acoustical ohms over the relatively small range of 300 to 700 cps.

The directional pattern produced by a horn is almost completely a function of mouth area and shape, and follows the same basic laws that determine directional patterns of cone-type units. Horn taper has only a minor effect. A horn designed for wide-range reproduction has a relatively large mouth, and as a result, will cause high frequencies to be beamed sharply. This effect can be eliminated by dividing the horn into a number of smaller horns supplied in parallel by the same driver unit. At the higher frequencies, each section of the multiple-section horn acts independently and provides a beam narrow enough to prevent interaction with its neighbors. As illustrated in Fig. 4, several narrow beams spread over a wide axis achieves a wide coverage because they add vectorially.

The shape of the horn mouth governs the pattern, in that wider sound distribution is obtained along the narrow axis of the mouth opening. This is why you see circular speakers mounted at ceiling level with the mouth pointed downward, and rectangular-shaped horns mounted with the widest mouth opening in a vertical direction. In either case, the idea is to get maximum area coverage.



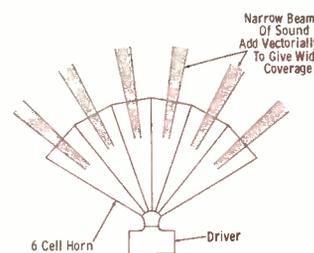
**Fig. 3. Horn-type speaker showing relative position of throat and diaphragm.**

#### Driver Units

If we consider the driver portion of a horn speaker as the sound generator, and visualize that it is in many ways very similar to a cone-type speaker, the driver unit will cease to be that mysterious metal thing on the rear end of a horn. While the power output of any horn is limited by its throat diameter (if a reasonable distortion figure is to be preserved), it is even more limited by the driver assembly. This factor is emphasized by the availability of hardware which permits two drivers to be coupled to a single horn for added power output.

Refer again to Fig. 3 and note that a basic driver consists of a magnet, voice coil, diaphragm, air chamber, and throat. There will be differences in the size, shape and movement of these parts and even additional components in certain cases, but the same basic configurations will apply, just as they do for cone-type speakers.

It was briefly mentioned earlier that a horn-type speaker was more efficient than a cone-type. Efficiency in a horn-type speaker is defined as the ratio of useful acoustical power output to electrical power input. Such factors as flux density, voice-



**Fig. 4. Narrow beams from multicell horn blend for wide dispersal of highs.**

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coil mass, diaphragm mass, ratio of diaphragm area to throat area, dimensions of the air chamber, and voice-coil temperature all affect efficiency. The air chamber is an important part of a driver unit designed for a wide frequency range because it provides the necessary transition between the diaphragm and throat areas which are of different sizes. In order to prevent distortion, the dimensions of the air are small in any direction when compared to the wavelengths of the higher frequencies. Further, if the driver is intended for high frequency use, say 10 kc and up, the distance from the throat opening to the farthest point on the diaphragm must be less than  $\frac{1}{2}$  wavelength at the highest frequency to prevent interference and radiation from the surrounding support or diaphragm suspension. This has led to the use of equalizing plugs between the driver throat and the horn itself. Instead of one circular opening, a plug provides additional passages between the diaphragm and the horn as illustrated in Fig. 5. All passages must be of equal length so that pressure variations will be in phase when they reach the throat of the horn.

The acoustical power output from a driver can be limited by the generation of spurious harmonics or subharmonics. These spurious generations can be caused by air overloads, excessive driving signal, varying voice-coil air-gap flux product, or nonfundamental vibration nodes of the diaphragm. This means that the maximum power rating for a horn speaker will be most often determined by the maximum tolerable distortion. Maximum permissible voice-coil temperature is another limiting factor, especially in high-frequency units.

Because the driver contains the moving portions of a horn-type speaker, it is the most likely source of distortion. Since diaphragm ex-

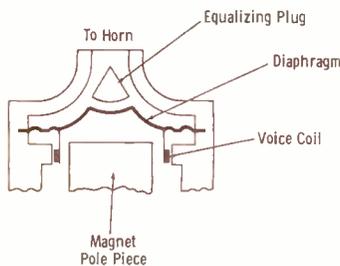


Fig. 5. Construction of horn driving unit showing location of equalizing plug.

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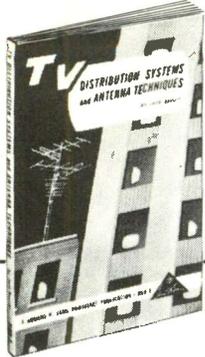
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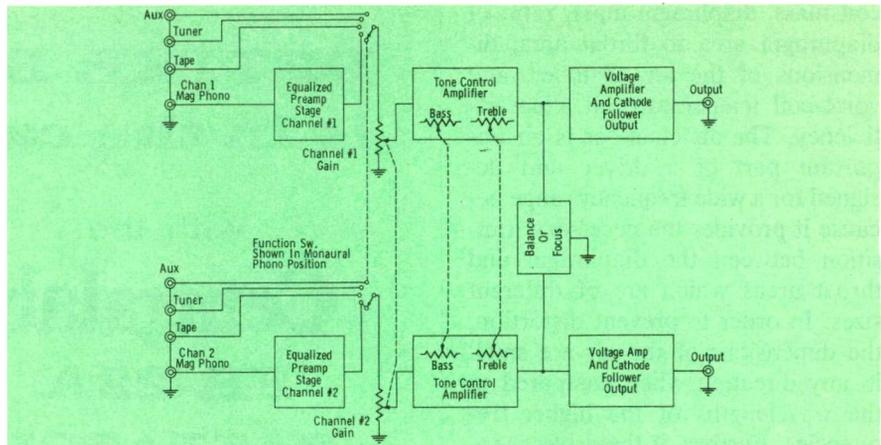


Fig. 6. Block diagram of basic stereo preamplifier showing pair of identical sound channels, ganged controls, balance network, and function-switch circuit.

curtion for constant power output is inversely proportional to frequency and the mechanical impedance of the system, the suspension system will cause the greatest amount of distortion at low frequencies. Considering this, it is apparent that a horn-type speaker should not be used in any manner other than the one for which it was designed. In other words, a low-frequency horn is good for low-frequency reproduction, a high-frequency unit for high frequencies, etc. Never apply large-amplitude, low-frequency signals to high-frequency horns (tweeters) or high-frequency signals to low-frequency horns (woofers). To do so is to invite disaster.

### A Glimpse into Stereo Preamp Design

To say a preamp is a preamp and a stereo preamp is simply two preamps would be like saying any two babies are twins. Since we know a 6SN7 won't work as the final amplifier in a 100-kw transmitter stage, it can be realized that a stereo preamp isn't just two of everything. While two of everything *is* required, there are several other considerations which are of utmost importance to stereo operation. In the block diagram of Fig. 6, for example, you will notice such things as balance control, ganged controls, monaural phono position and two input jacks for each position on the function selector (except monaural phono). Ganged equalization switches that simultaneously provide the desired equalization to both preamplifier stages are also often provided.

Of particular interest is the fact that the function selector switches,

gain, treble, and bass controls of the two channels are ganged—making this portion of the stereo preamplifier no more complicated to operate than a single channel preamplifier.

The gain controls in Fig. 6 might be labeled volume or loudness in a commercial unit. The function would be the same, however — to vary the output signal of the two channels equally and simultaneously. The ganged bass and treble controls need no explanation other than that they provide simultaneous changes in bass or treble response for the two channels. The design of the circuit in each channel must consider this ganged operation and the components selected must be of close tolerance to insure a balanced operating condition.

The balance or focus control is peculiar to stereo preamps — monaural units don't have one. The function of this control is to provide precise balancing of the two output signal amplitudes. If for any reason, such as unbalanced gain in the preceding stages, the output signals are not identical in amplitude, the balance control can be adjusted to achieve the desired degree of equality. While the balance control is shown connecting the two channels, it does so only from a DC bias standpoint. It does not link the signals in the two channels, or in any way cause the two signals to be mixed.

The sample circuit in Fig. 7 shows how the balance control connects into the circuit to do its required job. As you can see, the control is in the common grid return, and varies the ratio of DC resistance in each grid

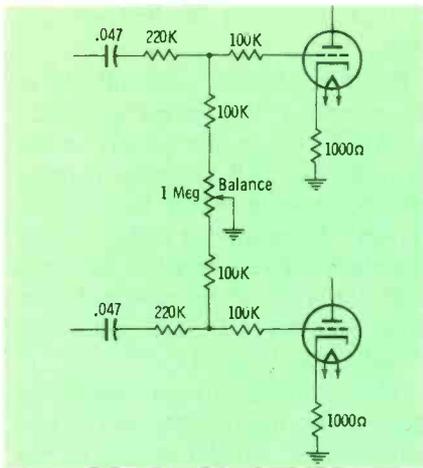


Fig. 7. Schematic of balance control network between stereo preamp channels.

circuit. Because of this feature, a stage of gain is provided after the balance control and just ahead of the usual cathode-follower output stage. This means that where some commercial preamp designs have employed only two dual-triode tubes (one tube as equalized preamp, 1/2 tube as tone-control amplifier and 1/2 tube as a cathode follower), many stereo preamps will employ three dual triodes in each channel. When this is done, one tube will probably be used for equalized preamp stages, one as a tone-control amplifier, and one as a voltage amplifier and cathode-follower output.

The major problem in stereo preamp design is construction of the unit so that it will provide identical performance on both channels without interaction between the two. This is achieved by the use of close-tolerance components, matched tubes, adequate filtering and decoupling between the two channels and the power supply, and careful layout of components so that ground loops will not be created. While these factors are most important on the manufacturing level, they also come into the picture whenever service is required. The serviceman should be careful to replace defective components with replacements having identical ratings and characteristics. Further, replacements should be installed exactly as were the originals; do not move ground connections, etc., even though it may be physically easier to connect a part differently. Also, any tube selected for use in a stereo preamp should be aged in the circuit for 24 hours before the unit undergoes a final check prior to its return. ▲



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# notes on TEST EQUIPMENT

*Informative reports from the lab*

by Les Deane

## New Transistor Checker

With the use of transistors continually increasing, the service technician needs a reasonably accurate means of evaluating their operating conditions. Having a transistor checker is imperative because one cannot merely look at a transistor and expect to see a lit filament, nor can he quickly substitute a replacement unit as he can with many vacuum tubes.

Featured in the photo of Fig. 1 is the Model 690 Transistor Tester recently produced by Triplett Electrical Instrument Co., Bluffton, Ohio. This battery-operated instrument is designed to measure leakage and gain characteristics of all low and medium power p-n-p and n-p-n type transistors.

Specifications and test features are:

1. *Power Requirements* — 1.5- and 30-volt self-contained batteries; universal replacements available.
2. *Panel Features* — leakage-gain meter; NPN/OFF/PNP selector switch; gain test button; calibration button and adjustment; one standard transistor socket and separate collector, base, emitter test jacks with external leads.
3. *Transistor Tests* — measures base-to-collector leakage current with open emitter element, current gain percent with grounded base element.
4. *Accuracy* — meter reading within 2% of full scale deflection on 0-500 DC  $\mu$ a range.

Checking a transistor usually involves two separate operations, similar to the shorts and emission tests performed on vacuum tubes; name-

ly, these are tests for leakage and gain.

Following instructions outlined in the manual, I first turned the CAL. ADJUST control completely counterclockwise. This prevents pinning the meter needle in case the unit under test is excessively leaky. I next consulted a transistor guide to determine the type of transistor I was about to check. In this particular case, it was an n-p-n.

After connecting the transistor to each of the appropriate test leads, I flipped the selector switch to the NPN position. I then pressed the CAL button and increased the CAL. ADJUST control until the meter needle coincided with the CAL. SET mark at full scale. Releasing the button, I was then able to observe the leakage current on the lower scale of the meter. In this instance, the needle barely moved away from zero, staying in the green area labeled GOOD. Incidentally, the green area of both meter scales is divided into two sections — one of dark green for small-signal transistors and one of light green for medium-power units.

Following this satisfactory leakage indication, I proceeded to the second step, which was a check of current gain. Without disturbing the CAL. ADJUST, I depressed the GAIN button and noted results on the upper scale. This scale is calibrated in percentage of gain based on full-scale deflection of 100. If the needle comes to rest above the

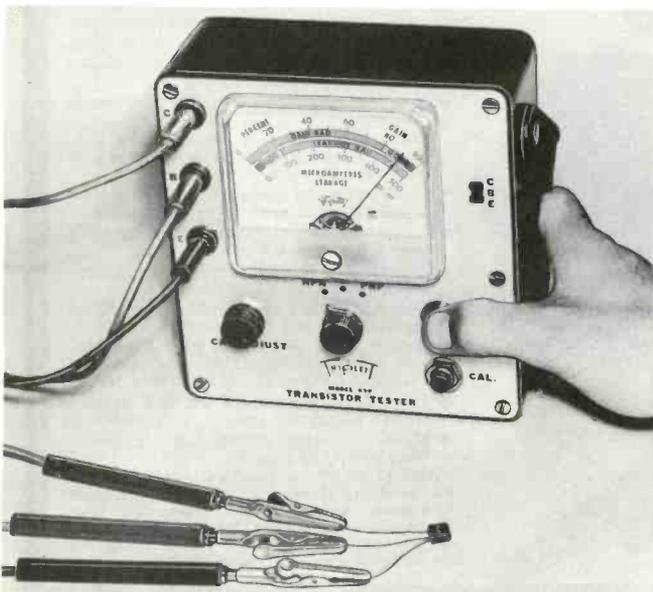


Fig. 1. New transistor checker by Triplett features panel test socket and detachable leads.

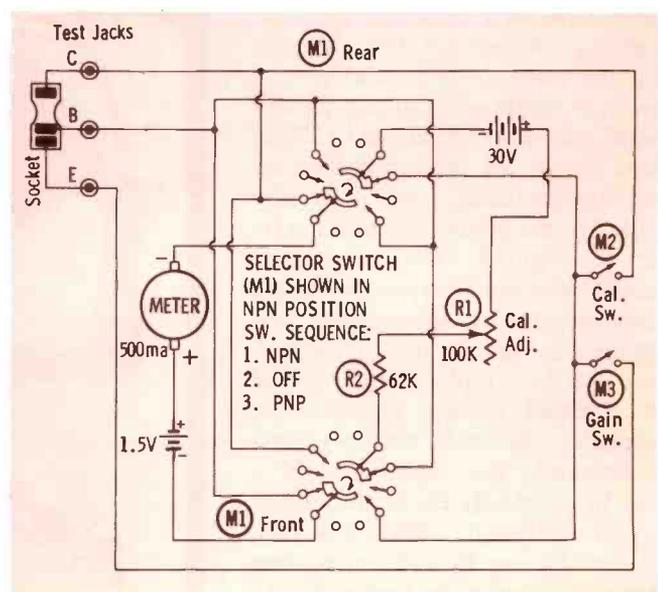
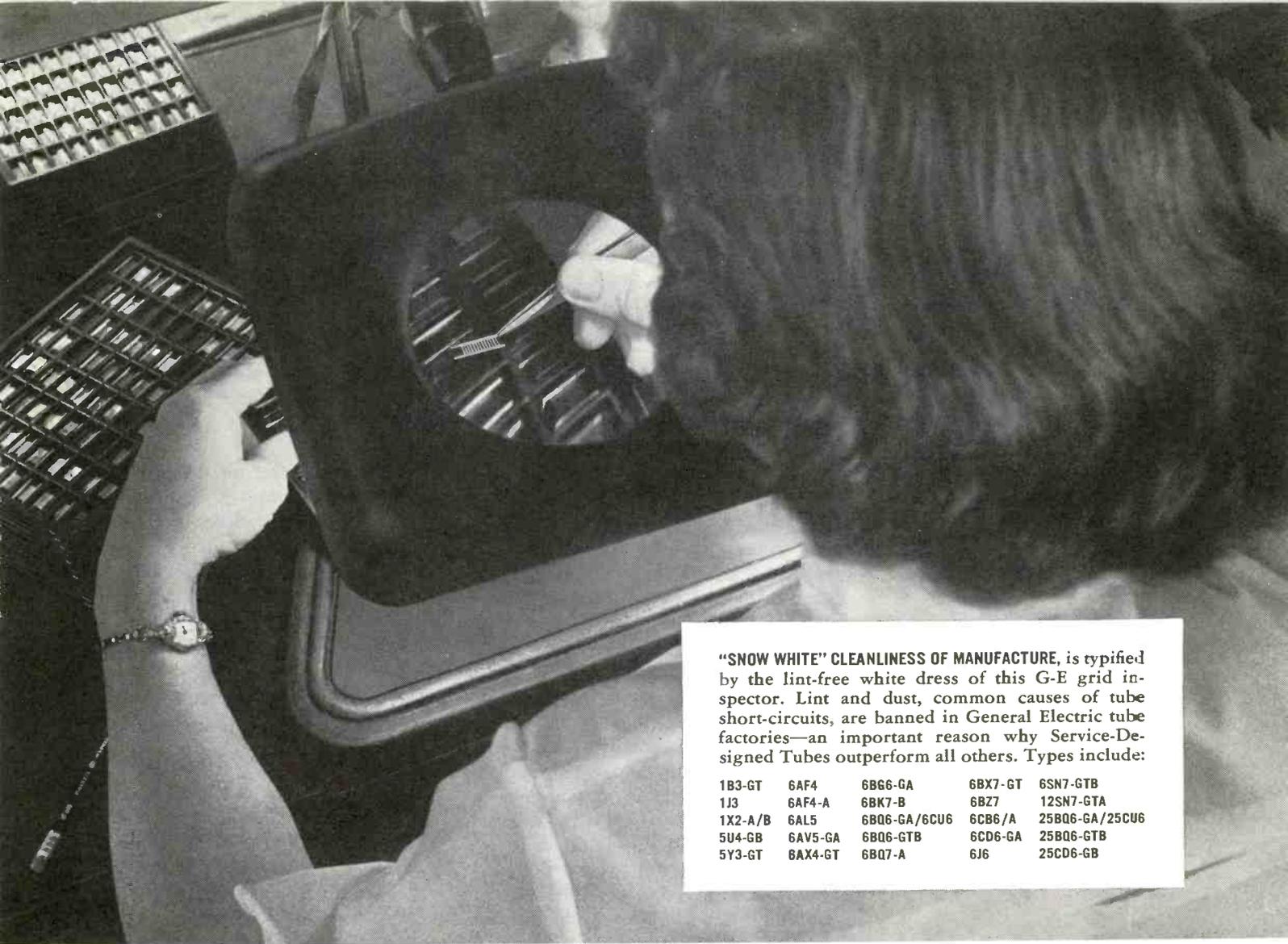


Fig. 2. Schematic of Model 690 shows meter and battery connections in various switch positions.



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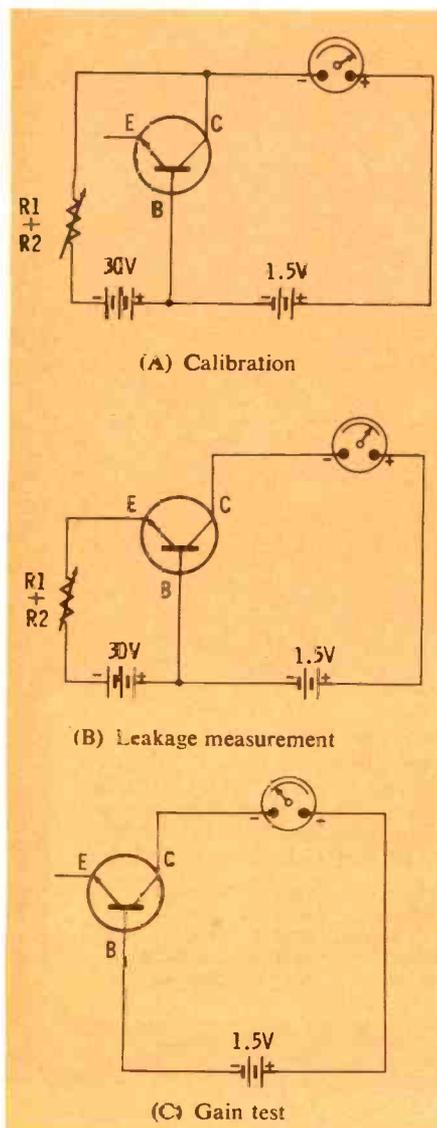


Fig. 3. Circuits for three-step operation of checking transistor in Model 690.

80% mark for medium-power transistors, the unit has sufficient gain. For small-signal units, however, the needle must register 90% or above. The particular transistor I tested produced a reading slightly above the 90% mark, indicating that its gain factor was such that the unit should function properly in its designed application.

So much for the mechanics of checking a transistor on the instrument. Now let's see how it actually does it. A complete schematic diagram of the Model 690 is given in Fig. 2. With the selector switch in the NPN position (as shown) and the CAL. switch (M2) closed, the transistor is first thrown into a circuit represented by the simplified schematic of Fig. 3A. The meter is then calibrated by adjusting R1 for full scale deflection.

After the calibration step, the CAL. switch is opened and the circuit loop containing the 30-volt battery is eliminated, leaving the simple arrangement shown in Fig. 3B. The current now flowing through the meter represents only base-to-collector leakage, which is indicated directly on a 0 to 500  $\mu$ a scale.

In the final test step, the GAIN switch (M3) is closed, thus connecting the 30-volt loop to the emitter (see Fig. 3C).

A portion of the emitter current must now flow in the base to provide transistor bias; therefore, the re-

maintaining current flowing in the collector (meter) circuit represents transistor gain. This, in turn, is indicated directly on the percent-gain scale of the instrument.

### Waveform Analyzer

The instrument pictured in Fig. 4 may look like a small portable scope with a simplified front-panel arrangement; actually, it's much more than that. Properly identified as the Model EA-1 Absorption Analyzer, the unit is manufactured by Kingston Electronic Corp., Medfield, Mass.

The EA-1 will permit the observation of waveforms in electronic equipment by either direct connection, or through use of a special electrostatic pick-up which is sensitive to signal radiations emanating through the glass of a vacuum tube.

Specifications and features are:

1. Power Requirements — 110/120 volts, 60-cps; power consumption 55 watts; line fuse provided.
2. Vertical Input — neutrode Standard Coil tuner converting all RF frequencies between 3 and 240 mc to a 43-mc IF (12 snap-in tuner strips are included with instrument); sensitivity of from 200 to 600  $\mu$ v/inch over a frequency range of 20 mc to 250 mc; direct frequency response down not more than 50% from 20 cps to 300 kc with a sensitivity of 2 mv/inch.
3. Horizontal Sweep Frequency — 5 ranges from 20 cps to 45 kc continuously variable; external sweep input provided with a frequency response down not more than 50% from 20 cps to 70 kc and a sensitivity of 500 mv/in, external sync input and amplitude control provided.
4. Front Panel Features — beam intensity control and on-off switch, sweep-frequency vernier, sweep range selector, vertical gain control, band selector switch, and input circuit selector; vertical input jack, ground jack, external sync input jack, external sweep input jack, and sound jack for monitoring signal by phones.
5. Rear Chassis Adjustments — focus, sync amplitude, sweep width, vertical and horizontal centering.

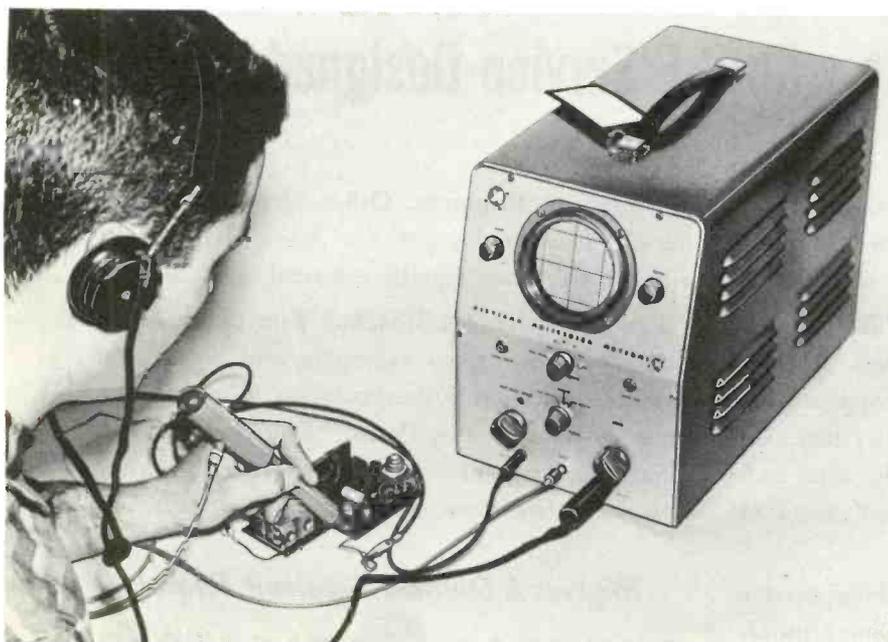


Fig. 4. Kingston Absorption Analyzer picks up signal radiation, converts it to standard IF frequency, and feeds it to high-gain scope for waveform analysis.

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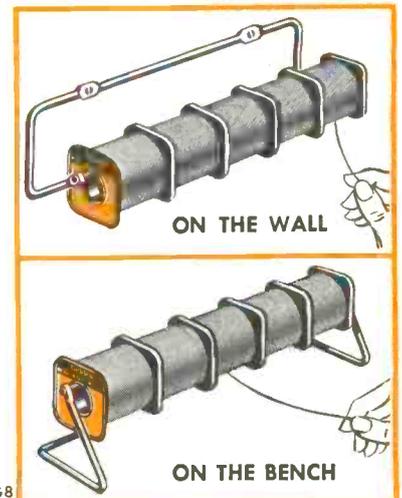
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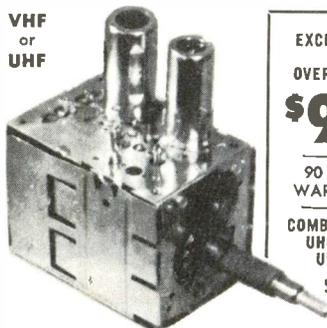
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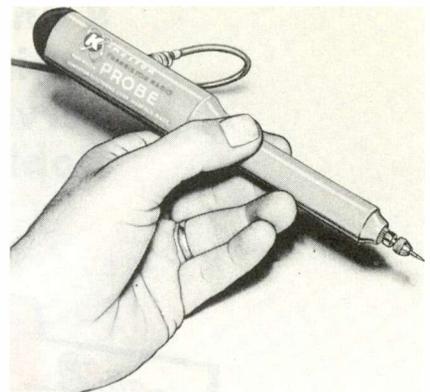
6. *Other Features* — two different sized ring pick-ups, one crescent probe, an envelope probe, coaxial cable and ground lead are supplied with the instrument; a direct probe with built-in step attenuator, a transistor radio probe, and a special envelope probe for 7-pin miniature tubes are also available as accessories.

Examining the Model EA-1 in the lab, I noted that it has all of the potentials for TV troubleshooting possessed by its predecessor, the Model VS4 (see this column in the July, 1957 issue). In addition to the many servicing features of the VS4, the new Analyzer can be used to analyze just about any piece of equipment having periodic or recurrent waveforms. Equipment for two-way communications, commercial or governmental aircraft, navigational or weather radar, public address systems, tape recorders, record players, hi-fi systems, electronic organs and counters all fall into this category.

I obtained one of the Transistor Radio Probes (Fig. 5) with the Model EA-1, and thought you would like to see how this accessory item is used in conjunction with the Analyzer to troubleshoot a typical transistor radio.

Following instructions, I first connected the probe to the Analyzer's input cable and attached the ground lead to the radio chassis under test. Next, I placed the input selector in the VERT. AMP position, meaning that the input signal would be applied directly to the vertical deflection system of the scope. Since I not only wanted to obtain the signal waveform but also an audible indication, I secured a set of ear phones and plugged them into the sound jack located in the lower-left section of the front panel.

Adjusting the gain to near maximum, I then turned the instrument on. After it had warmed up a moment, I touched the probe tip to the collector element of the RF transistor and varied the tuning gang until I picked up a strong local station. If the antenna input or RF stage had been defective, I would have not received a signal. I noted that weaker signals could also be picked up at this point by increasing Analyzer gain to maximum and placing my hand close to the receiver antenna.



**Fig. 5. Detector probe used with Model EA-1 to signal-trace transistor radios.**

Progressing from one stage to another, I found that I could signal trace the entire radio by adjusting Analyzer gain for convenient viewing and listening level. Although ear phones are not always required in this operation, they come in handy when troubleshooting for a distorted condition. I found too that the sound may distort if the Analyzer gain is excessive; therefore, it may be necessary to reduce the setting as you proceed toward the output section of the receiver.

**Automatic Tube Tester**

What's an automatic tube tester? Well, the instrument shown in operation in Fig. 6 is a dynamic mutual conductance tester which approaches automation by instantly setting up various tube-test conditions with the simple insertion of a prepunched index card.

Designed especially for the serviceman, this new "Cardmatic" Model 121 is manufactured by Hickok Electrical Instrument Co. of Cleveland, Ohio.

*Specifications and test features are:*

1. *Power Requirements* — 105/125v AC, 50/60-cps; line voltage adjustment and overload indicator provided.
2. *Tube Tests*—leakage and shorts, quality (mutual conductance), gas and grid emission with built-in VTVM; twin-section tubes are tested automatically by pushing a button; special application tests provided for certain tubes.
3. *Panel Sockets* — 10 different types provided with 320 specific heater voltages available.
4. *Other Features* — space for over 600 index cards (approximately 325 furnished); detachable lid; case size and weight, 19¾" ×



Fig. 6. Hickok Model 121 "Cardmatic" is a dynamic mutual conductance tube tester which uses prepunched, coded cards to set up conditions automatically.

13 1/4" x 8", 34 lbs.

When checking out this piece of equipment, I was first concerned with testing a number of tubes to see how automatic the operation really was, and how speedily a tube could be checked. The thing that impressed me was that the "Cardmatic" is a dynamic type checker, yet can be set up as quickly, if not more so, than any other type on the market.

It actually takes longer to tell about it than to do it, but to familiarize you with the operation of the instrument, here are the detailed steps involved.

One unusual thing I noticed during the preliminary set-up was the three-prong "Underwriters - Type" plug on the end of the power cord. This is an added safety feature where the third prong connects to ground through the AC socket or a

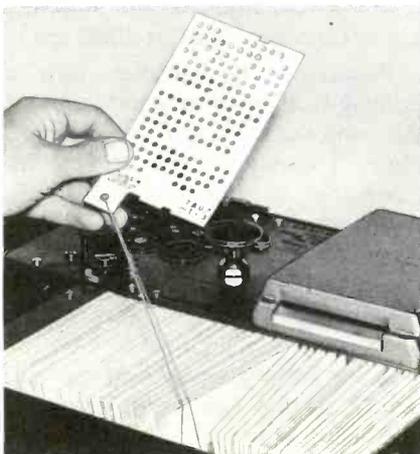


Fig. 7. Cards for Model 121 are on a cord and can be quickly returned to file.

socket adapter. After properly connecting the cord to a power source, I turned the instrument on by rotating the LINE ADJUST knob from its OFF position to the 115-volt point. Permitting the unit to warm up for about 25 seconds, I then zeroed the meter (VTVM) using the control provided on the front panel. With these steps completed, the Model 121 was now ready to check tubes.

Selecting a card from the index rack which corresponded to the tube I was checking, I inserted the tube into the correct socket and pushed the card into the automatic switch. Giving the tube about 5 seconds to warm up, I then pushed button #1 and set the line voltage so that the meter needle registered on the LINE TEST point. Releasing the button, I glanced at the 5 short lamps located between the meter and the switch. A short was not indicated, nor did any leakage show up on the meter, so I proceeded to push button #2. This is the quality test, and the condition of the tube is indicated immediately on the GOOD — REPLACE scale.

As a final test, I pushed button #3 which checks the gas content in amplifier tubes. Referring to the meter, I found that the needle remained in the green area marked GOOD on the gas scale. In my first few testing attempts, I was a little slow — for it took me a little less than one minute to complete the preceding test steps. After I be-

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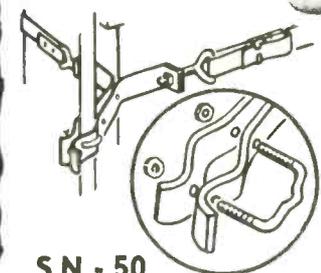
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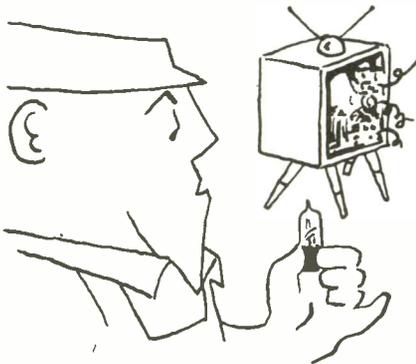


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came familiar with the instrument's card filing system, however, I found that I could check a typical IF tube (such as a 6CB6) in about 15 seconds. On one occasion, it took me only 6 seconds to discover a short in a 6W4 damper tube, so I'm sure that with experience one could learn to test a tube in the 8 to 12 seconds specified by the manufacturer.

The card filing system is shown in the photograph of Fig. 7. Note that the test cards are all attached to a single nylon cord. As a card is pulled from the tray, the cord also extends outward, being anchored to a reel mechanism similar to a window-shade roller. When a test is completed, the card is removed from the switch and is more or less free to follow the attached cord back to its original position in the tray. Detailed instructions for adding or removing test cards are given in the "Cardmatic" manual.

### Scope Features Portability

The oscilloscope pictured in Fig. 8 is a product of Simpson Electric Co., Chicago. Designed for general-purpose applications, the Model 466 "Handiscope" features a 5-inch CRT with green calibrated mask. The instrument is suitable for just about any TV or radio troubleshooting job, and due to its size is easily transported.

Specification features are:

1. **Power Requirements** — 105/125 volts, 50/60 cps; power consumption 50 watts; line fuse provided.
2. **Vertical Input System** — frequency response 15 cps to 100 kc  $\pm 1$ db; down 6db at 250 kc, useful to 1 mc; sensitivity of .03 volts RMS/inch; maximum input voltage 400 volts p-p; input impedance on  $\times 1$  position is 100K ohms shunted by 40 mmf; on  $\times 100$  position 500K ohms shunted by 35 mmf; input cable supplied.
3. **Horizontal Input System** — frequency response 15 cps to 20 kc  $\pm 1$  db, down 6 db at 100 kc; sensitivity .7 volts RMS/inch; maximum input voltage 150 volts p-p; input impedance 250K ohms shunted by 40 mmf.
4. **Sweep Frequency System** — internal range of from 15 cps to 80 kc in six steps; external sweep position and variable



Fig. 8. Portability is main feature of Simpson 466 general-purpose scope.

phase provided.

5. **Synchronization System** — positive or negative sync voltage from vertical system or from external source.
6. **Z-Axis Modulation** — with normal intensity, approximately 20 volts RMS required to extinguish beam; input provided on rear panel.
7. **Voltage Calibration** — special jack provided on front panel; 1-volt peak-to-peak  $\pm 10\%$ .
8. **Size and Weight** — overall 7 $\frac{7}{8}$ "  $\times$  12 $\frac{1}{4}$ "  $\times$  16 $\frac{1}{8}$ ", 19 lbs. net.

Working on some audio equipment in the lab the other day, I used the Handiscope to check overall operating characteristics of a commercial amplifier chassis. Arranging the apparatus as illustrated in Fig. 9, I connected an audio generator to the input of the amplifier and set the frequency at 1000 cps.

Replacing the speaker with a dummy-load resistor which equaled the voice-coil impedance of 16 ohms, I connected the vertical input cable of the Handiscope directly across the resistive substitute. Naturally, I employed a resistor having a wattage value equal to or greater than the output power rating of the amplifier.

After turning everything on and permitting the equipment to warm up for a few minutes, I set the volume control of the amplifier to maximum and adjusted the sweep

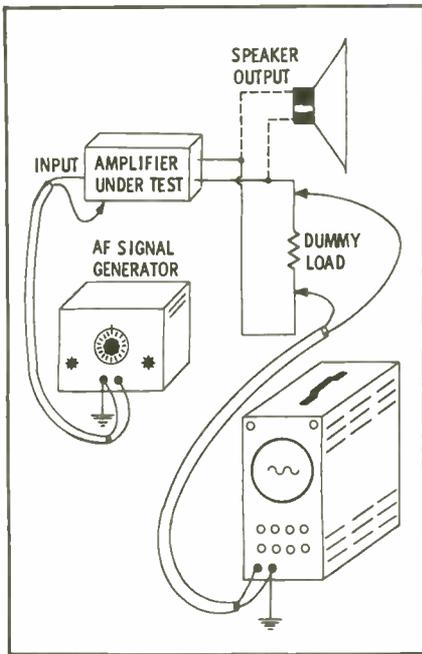


Fig. 9. How to use the "Handscope" to check for audio amplifier distortion.

of the scope to 500 cps—(half the repetition rate of the signal to be viewed so that two complete cycles of the sine wave would be reproduced). Next, I set the signal level of the generator slightly below the point which produced clipping of the sine wave peaks in the output. Under these conditions, the power dissipated in the load should equal approximately the rated power of the amplifier.

To check this point, I measured the AC output voltage across the dummy load and found it to be about 12.5 volts. Knowing that  $E^2/R$  equals power, I calculated the output to be roughly 9.8 watts. Since the amplifier under test was rated at 10 watts, this value seemed to be fairly satisfactory. Oftentimes you may find this power figure to be considerably lower than the rated value, or the shape of the output wave might look distorted. This, of course, would indicate trouble in the amplifier and would call for additional troubleshooting procedures. I found, for example, that changing the bias of a stage will usually result in unequal peaks within each cycle. An overload, on the other hand, will generally produce a flattening or clipping of peaks on both halves of the wave. The manual for the Simpson scope outlines several other interesting applications for the instrument and also contains servicing hints, schematic, and complete parts list. ▲

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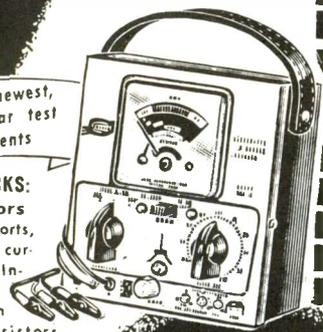
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(Continued from page 8)

quoted fee of \$70. (Consider for a moment what might have been your feelings under similar circumstances. If you asked yourself the question, "Do I do, or do I don't?" I hope your resolve was, and forever is, *do!* For only by doing are you learning, and learning promotes that great confidence and ability so necessary to a successful servicing business.)

**Action**

No, the solution wasn't easy. In order to establish customer confidence, it was necessary to exhibit on-the-spot technical skill. And to make sure the job was complete in all respects (there were evidences of fresh soldering at a number of connecting junctions), the receiver would most certainly have to be taken into the shop for the actual repair. In this instance, the receiver was removed from its cabinet and the speaker (a PM without mounted audio transformer) was disconnected. Next, the low-voltage rectifier tube was taken from its socket, and a quick continuity check was made from the 5U4-G socket terminals 2 and 8 to chassis ground. The reading was better than 20,000 ohms, which indicated that the filter capacitors were not shorted. Power was supplied to the receiver and AC voltage measurements were taken at pins 4 and 6 of the 5U4-G socket. The transformer was also felt for any evidence of excessive heat. At both points, a cool 320 volts AC registered on the meter, exactly the value listed for this particular receiver.

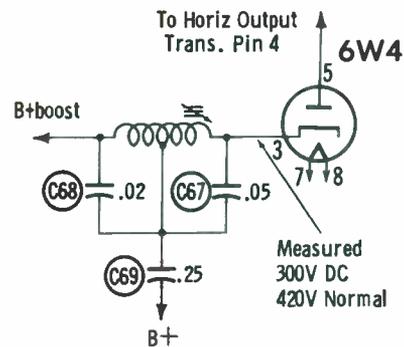
**Results**

The power transformer shorted? Hardly, under these excellent conditions. The customer was so informed and also told that since the short (with the rectifier tube in the circuit) would undoubtedly be difficult to find, the work should be performed in the shop where test aids and equipment were readily available. To this the customer promptly agreed and asked also that the picture tube be checked when the other repairs had been completed. During this time, while I had been making tests, there had been no mention of price. Neither was anything said now—a silent, mutual understanding had been reached. In the eyes of

the customer, I was competent and the bill would be paid.

**In the Shop**

Needless to say, this is where the fun began. A new 5U4-G was inserted in the tube socket and power was applied. The picture tube lighted momentarily and then the rectifier tube plates began to turn metallic red. The AC was immediately disconnected and a few futile resistance checks were made—no trouble was evident. Now this in itself was a substantial indication, for if no capacitor or transformer was shorting to ground, the only other probability was a tube that, when fired to full potential, would short internally. (All receiving tubes had, of course, been checked with a tester while the set was still in the home, but no shorts were discovered.) A quick look at the schematic seemed to implicate the horizontal output, horizontal multivibrator, damper, audio, and perhaps even the vertical output tubes. How to find the bad one was the problem. I could substitute all of them and come up with a quantitative but not qualitative solution. Since there was some limit on the price (and the CRT had not yet been checked), the qualitative approach was decided upon. The horizontal output tube was replaced first, but the rectifier tube still glowed red. Next, the 6V6 audio amplifier was substituted for; again, the low-voltage rectifier plates glowed angrily, answering my suspicions in the negative. Just by chance, I noticed the damper-tube plate exhibiting the same tell-tale red. Ah, here was the culprit at last. And sure enough it was, for substitution of this tube returned B+ voltages to normal, and the cathode-ray tube lighted again. The big problem had been licked—happy day! But hold up a moment—what about the little



**Fig. 2. Voltage on damper cathode measured 300 VDC instead of 420V.**

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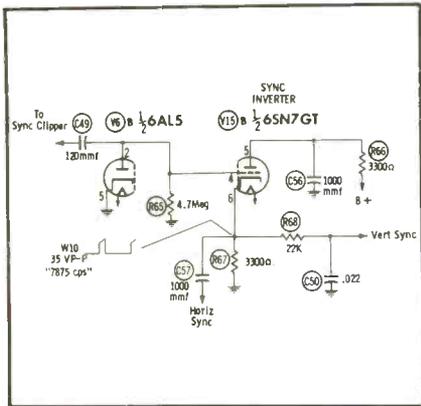


Fig. 3. W10 was incorrect because R67 was improperly wired into the circuit.

problems? And this is where so many television repair shops make their human, but fundamental, mistake. Hurray, they say, it works! But does it? Read on . . .

### Much, Much More

What about the smoking condition observed when the receiver was allowed to run in the home? Isn't smoke usually the result of excessive current through a resistor? Notice in Fig. 1 that focus coil T6 and its associated control (R5, 6000 ohms) act as the B+ choke. The element in this potentiometer had been charred, indicated by the rough feeling which was evident when its shaft was rotated. R5 was replaced with an equivalent-value wire-wound control.

Now what about the B+ boost circuit? Could there be any defects that caused the damper tube to go bad? The voltage at pin 3 of the damper (Fig. 2) was down to 300 volts. More tubes? Hardly, the previous tube check was probably reliable. More likely, this fault could be attributed to the capacitors associated with the development of boost voltage. One of the three capacitors was found improperly connected and another, when placed across a capacitance checker and subjected to a potential of 500 volts, proved to be leaky. Just to be safe, all three (.02, .05 and .25 mfd) were replaced and boost returned to a normal 420 volts.

Next, the waveforms for sync tubes V6B, V14 and V15B were checked for shape and amplitude. Essentially, they were within tolerable limits—all, that is, except the 35-volt positive - spike waveform

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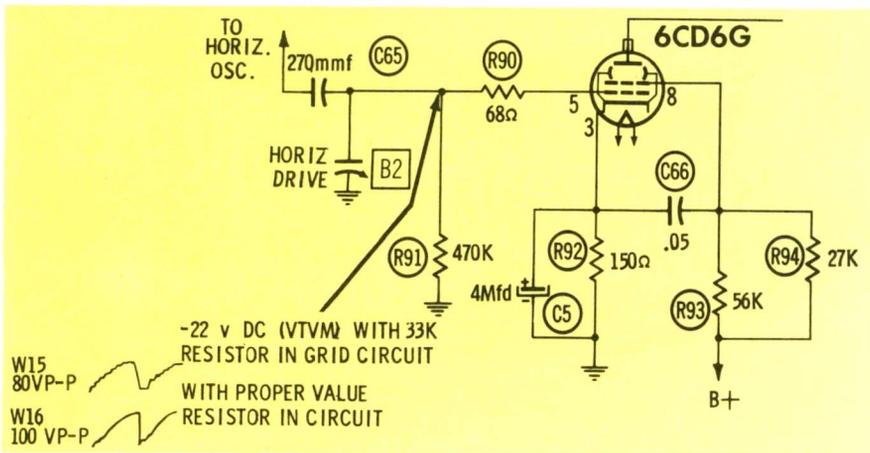


Fig. 4. Distortion of 6CD6 grid waveform was traced to incorrect value for R90.

W10 at the cathode of V15B in Fig. 3. Here, it was found that R67 had been connected *into* the circuit instead of from cathode to ground. By changing the circuit to conform with the diagram, the waveform specifications were satisfied.

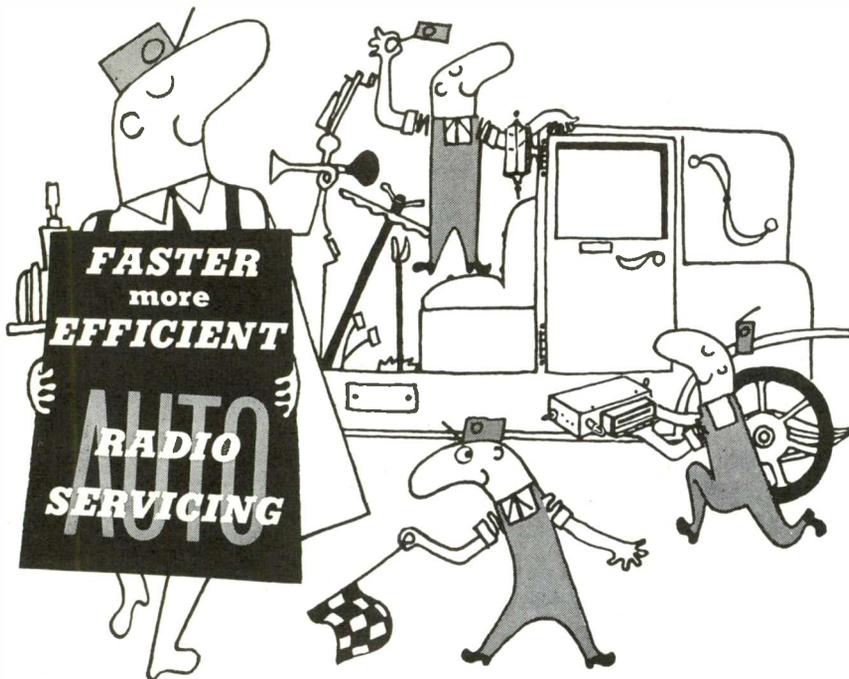
Making a further check, the peak voltage at the grid of V19, the horizontal output amplifier, was measured. Here, the waveform amplitude was essentially correct, yet the shape was considerably distorted. The top was unduly flattened (W15 in Fig. 4) and there were transients along the leading edge. A glance at the input grid resistor provided the answer. The color code was three orange rings. No wonder the poor waveform appeared out of shape—the value for R90 (which is always low) given by the manufacturer was a mere 68 ohms, not the huge 33,000 ohms that was in the circuit. The resistor was immediately replaced, and not only did the wave-shape return to a modified sawtooth (W16 in Fig. 4), but it also increased in amplitude by 20 volts! Three weak tubes in the video IF strip (discovered during tests made in the home) were next replaced and the waveform at the plate of the video output amplifier came up from 60 volts to slightly more than 80 volts peak-to-peak.

Now, it was time to check the picture tube for a hint of the distortion that had been noted at the plate of the 6AC7 video amplifier. A CRT adapter used in conjunction with the tube tester quickly confirmed the suspicion that the picture tube was shorted.

#### Back to the Customer

After several additional voltage and oscilloscope waveform checks had been made to firmly resolve the satisfactory over-all operation of the receiver, the customer was called. All circuit defects and their repairs were carefully reported before news of the defective cathode ray tube was gently broken. The customer "never even blinked" as the saying goes.

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## Troubleshooting Transistor Radios

(Continued from page 15)

tuning capacitor (Fig. 6) are generally equal in size and capacity. The latter is of particular importance since it makes differentiation between RF and oscillator sections a bit more difficult. The loop is always a handy guide in this respect because one end always connects to the RF section of the tuning gang. This will keep you from turning the wrong trimmers during the alignment procedure. The addition of the adjustable oscillator coil complicates tracking adjustments slightly because the coil must be set at the low end and the oscillator trimmer at the high end. This procedure must then be repeated at least twice to achieve optimum oscillator tracking.

### RF Tracking

Alignment instructions normally specify that the RF trimmer should be adjusted at approximately 1500 kc, which usually results in acceptable tracking of the RF section. Generally, however, improvements in performance can be made even after it has been aligned as prescribed. Remember that transistor radios, like everything else, are mass-produced—a factor considered in their design. Here is a check which quickly indicates whether or not further improvement is possible: Bring your hand near the loop antenna of the receiver; if the desired sound increases in volume, improvement can be made. If the noise level goes up, but the desired sound remains steady, no improvement can be made.

If sound level increases, and you

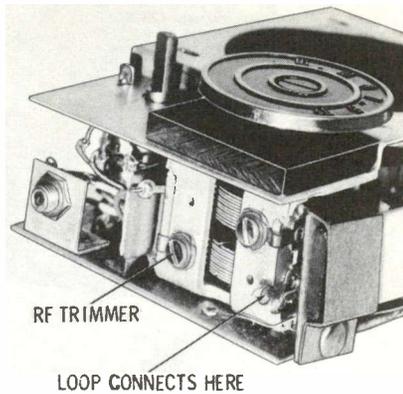
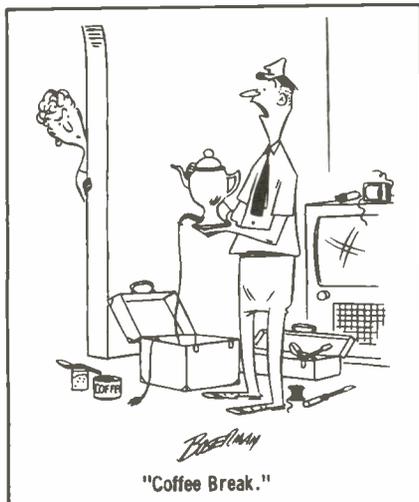


Fig. 6. Loop antenna hooked to tuning gang helps to identify the RF section.

wish to improve performance, unwind a couple of turns from the end of the larger loop winding (Fig. 7) and reverse-wind one turn. Set the signal generator frequency and the station indicator to about 1100 kc and adjust the RF trimmer to produce maximum VTVM indication. Repeat the hand test with the receiver tuned to a local station near the high end of the band. If maximum performance is now indicated, remove the reversed turn and an additional turn. After cutting off all excess wire, resolder the shortened coil to the tuning capacitor. Retune

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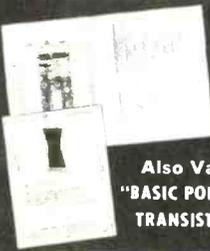
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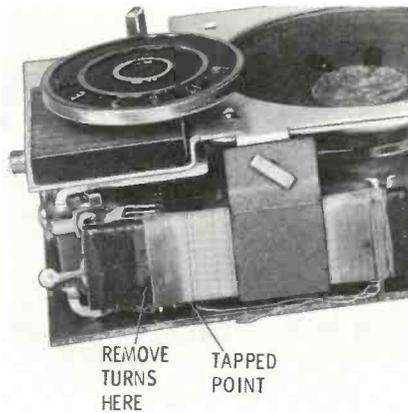


Fig. 7. Removing turns from this end of antenna loop helps improve RF tracking.

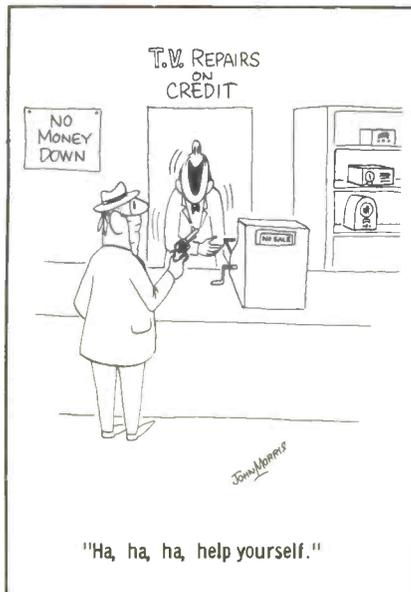
the RF trimmer to 1100 kc as before and repeat the hand check. Continue this process until optimum performance is indicated.

When optimum RF tracking is obtained, the RF trimmer capacitor will be about one-quarter turn from tight. Continue to remove turns from the loop and readjust the trimmer until this condition has been achieved. Once you have become familiar with this procedure, it can be completed in about 10 minutes. The improved performance of the radio will amaze you and overwhelm your customer.

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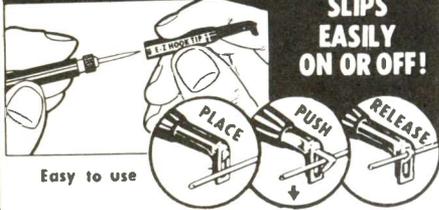
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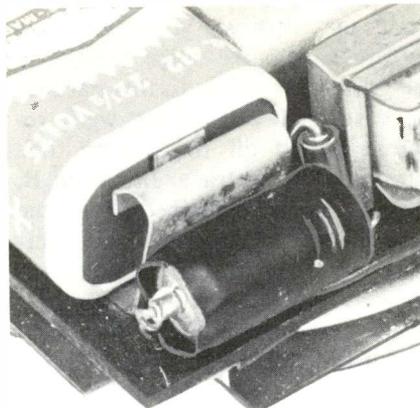


Fig. 8. Use of plastic tape to insulate the bodies of electrolytic replacements.

in most transistor radios, only resistors, electrolytics (single section), loop antennas and batteries are considered to be standard replacement items. You should have exact replacements or identical substitute units for tuning capacitors, volume controls, IF transformers, oscillator transformers, transistors, and audio transformers.

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## Vertical Retrace

(Continued from page 11)

transformer secondary (or the vertical yoke windings) may be of either polarity, depending on receiver design. Usually, the polarity of the secondary winding is not indicated schematically, and if unknown, a scope may have to be employed to determine which it is.

### Vertical Blanking Circuits

Several typical retrace elimination circuits are shown in Fig. 6. In circuit A, there is a direct connection

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from the vertical yoke windings (where a negative spike exists) to the grid of the CRT. In this particular case, 100-volt pulses were required to cut off the picture tube during vertical retrace. In circuit B, a spike is also obtained from the yoke, except that here an RC network is used instead of a direct connection. The schematic of Fig. 6C shows a circuit employing a vertical output transformer of the autotransformer type, with a capacitive-divider network to supply a positive spike from a tap on the transformer to the picture tube cathode.

In circuit D, blanking pulses are taken from the plate of the vertical oscillator tube. This blanking circuit happens to be part of the sawtooth-forming network of the oscillator. Its circuit explanation is as follows: C56 is the sawtooth-forming capacitor which provides a signal to drive the vertical output tube. During the vertical oscillator quiescent period (vertical trace time), C56 charges with the indicated polarity through resistors R4, R42, R43 and R44. The positive voltage drop across R43 does not cut off the picture tube beam, which is exactly what is desired during vertical trace time.

When the vertical oscillator conducts (during retrace time), C56 discharges through R43, R44, the vertical oscillator tube, and the primary of T1. The polarity across R43 is thus reversed, biasing the picture tube sufficiently to effect beam cutoff. (Note that the voltage produced is a 30-volt negative spike at the picture tube grid.)

In Fig. 6E, the vertical blanking circuit consists of several RC components and a vertical blanking tube. The tube is an ordinary triode amplifier which serves to amplify the blanking pulse before it is applied to the CRT grid. Ordinarily, the vertical output signal (or the high-side of the vertical output transformer primary) connects to the picture tube cathode (Fig. 5) since its spikes are positive; however, because the vertical blanking amplifier reverses the polarity of the spikes, its output is fed to the picture tube grid.

Example F in Fig. 6 illustrates the vertical blanking circuit used in a color set which, you'll notice, also employs a horizontal blanking circuit. Both circuits are cathode fol-

lowers. In the vertical blanking circuit, the high side of the vertical output transformer primary is fed to the grid of a blanking tube and the output is taken from its cathode. The cathode follower is used to prevent loading of the video amplifier. Because the output signal of the cathode follower is not reversed in phase from its input, the take-off voltage is fed to the picture tube cathode.

It was mentioned earlier that the vertical blanking circuit usually connects to the undriven picture tube element in order to minimize video amplifier loading. Nevertheless, the vertical blanking circuit of Fig. 6G and the video amplifier both feed the three cathodes of the color picture tube. Isolation between the two circuits is maintained, however, by use of isolation transformer L23.

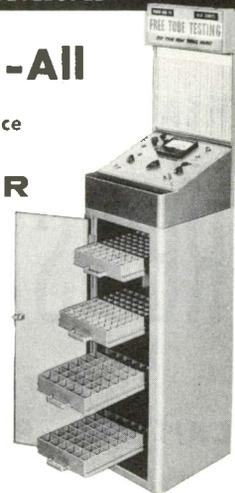
### Installing Blanking Circuits

If a receiver displaying retrace lines does not employ a blanking circuit and a fault in the video output or picture tube circuits is not the cause, any of the circuits shown in Fig. 6 may be used. It may be nec-

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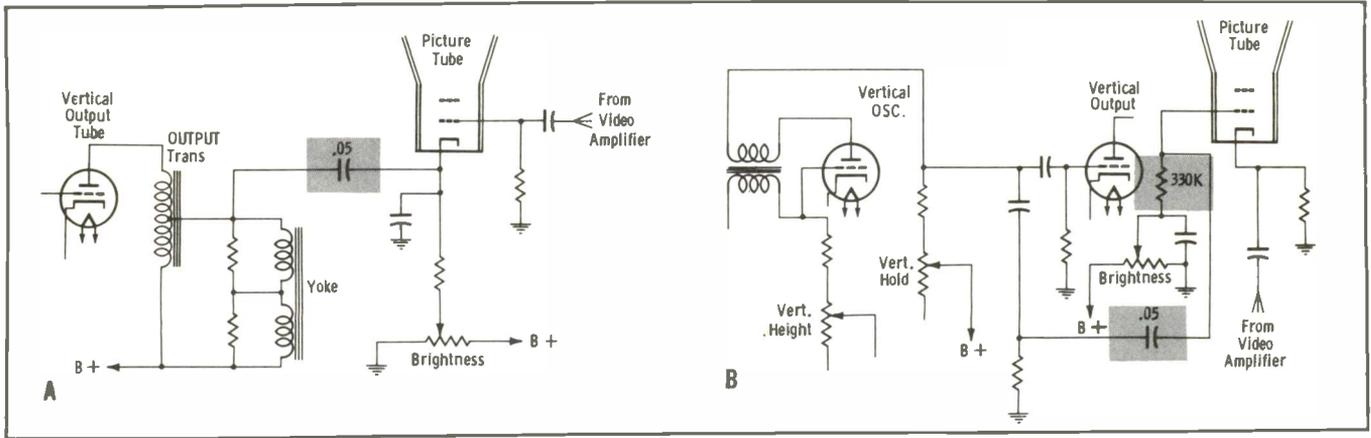


Fig. 7. Blanking pulse networks for picture tube cathode and picture tube grid.

essary in some cases to alter the values of resistors and capacitors in the circuit to conform with picture tube operating characteristics and p-p signal voltage in the vertical sweep circuit.

Fig. 7A shows a simple way to install a vertical blanking circuit. Simply connect a 0.05-mfd capacitor from the tap on the vertical output transformer or from the output tube plate directly to the picture tube cathode. If the peak-to-peak voltage of the spike is too high (say, over 100 volts), insert a 100K- to 220K-ohm resistor in series with the capacitor. If the connection is made to the secondary of an isolation-type transformer, then use a 10K-ohm resistor.

When the video amplifier drives the picture tube cathode instead of the grid, connect the 0.05-mfd capacitor as shown in Fig. 7B. In this circuit, the coupling capacitor connects from the junction of C1 and R1 of the sawtooth-forming network to the picture tube grid. If the picture tube does not contain a large-value resistor between the brightness control and grid, then insert a 330K resistor as shown. Otherwise, you may find the blanking signal being shunted to ground through B+ filters or a grid-bypass capacitor.

When choosing a point in the vertical sweep circuit from which to obtain a spike, try to use a voltage of about 30 to 50 volts p-p. Sometimes, however, a pulse voltage as high as 100 volts may be needed for adequate blanking. Check waveforms of the circuit to determine the peak-to-peak values if they are not known. When higher than needed, install a divider network similar to those shown in Fig. 6. ▲

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# PRODUCT REPORT

### Replacement Flybacks

Rogers has announced three new exact replacement flyback transformers for Bendix and Capehart TV sets. EFR 134 is equivalent to Bendix part number 265075-1, EFR 135 is a replacement for Bendix part 265078-3, and EFR 143 replaces Capehart part 850274B-1.



For further information, check 36U on Literature Card.

### Indoor Antenna

Radion has announced a novel indoor antenna, the X-5, which consists of a four-conductor lead wire terminated in a rotary switch. A matching network in the switch housing is adjusted in order to vary the electrical length of the wire and thus adapt the antenna to various reception conditions. List price is \$4.95.



For further information, check 37U on Literature Card.

### Self-Service Tube Tester

Vis-U-All is supplying a self-service tube tester to radio-TV service dealers. A floor model with tube storage cabinet (V11) and a counter model (V12) are both available. Either one occupies an area of 15" x 15". Revised tube charts are periodically provided by the manufacturer.



For further information, check 38U on Literature Card.

### Test Leads

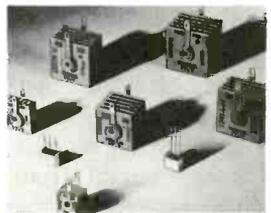
Alpha has designed a display rack for merchandising test leads. Each pair of leads is individually packaged in a plastic tube, and an assortment of 12 different combinations of test prods and meter tips is available. All leads are 50" long, contain a No. 18-gauge conductor, and have high-dielectric rubber insulation.



For further information, check 39U on Literature Card.

### Selenium Rectifiers

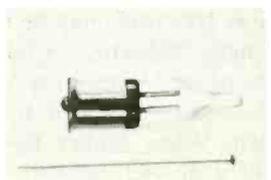
IT & T Components Div. has announced a new line of Federal selenium rectifiers known as the "Golden Eight." Two styles of potted dual diodes are available for horizontal AFC applications, and there are six different B+ rectifiers with ratings that range from 65 to 600 ma at 130 volts.



For further information, check 40U on Literature Card.

### Neon Lamps

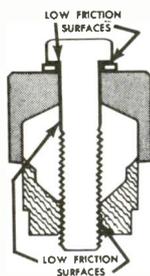
Circon has developed a new subminiature neon lamp, Type NE2R, for indicator, panel light, and computer applications. No external current-limiting resistor is required, since the necessary current-control network is built in. The standard version is for 110/125 VAC operation.



For further information, check 41U on Literature Card.

### Chassis Punches

Walsco has introduced a new line of "L. T." (low-torque) chassis punches. A new "electro-coating" process, applied to the steel used in these punches, is claimed to reduce friction between parts so that 50% less torque is required to cut through metal. Available in the new line are all standard sizes of punches needed for cutting round, square, key, and "D" holes in chassis.



For further information, check 42U on Literature Card.

### Miniature Multimeter

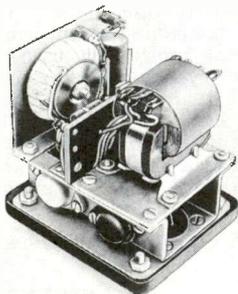
British Industries Corp. is importing the English-made AVO "Multiminor," a miniaturized volt-ohm-milliammeter that measures DC voltages from 2 mv to 1 kv at 10K ohms per volt; AC voltages from 200 mv to 1 kv; DC current from 2 ua to 1 amp.; and resistance from 5 ohms to 2 megohms.



For further information, check 43U on Literature Card.

### Transistorized Inverters

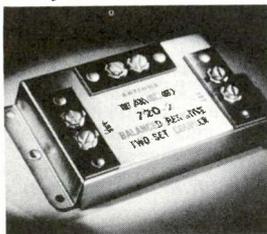
Universal Transistor Products Corp. is marketing an extensive line of completely transistorized DC-to-AC static inverters featuring extremely small size and light weight. Some units develop 115 VAC at 60 cps from a 6- or 12-volt source and therefore would be suitable for operating test equipment or small P. A. systems from an auto battery. Others have 24- to 28-volt input for mobile and aircraft applications.



For further information, check 44U on Literature Card.

### Multi-Set Couplers

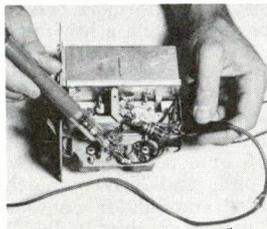
TACO has introduced a new line of balanced resistive couplers for connecting two or more TV sets to the same antenna with minimum interference between the receivers. Indoor two-, three-, and four-set types have list prices of \$2.95, \$3.50 and \$4.25 respectively. Also available are weatherproof two- and three-set couplers.



For further information, check 45U on Literature Card.

### Circuit Tracer

Hunter Tool Co. is supplying a pocket-sized "Circuitracer" probe for general circuit testing. A lamp in the body of the probe glows to indicate the presence of voltage across the two terminals of the tester. By adding a battery in series with the lamp, the user can trace continuity in circuits with the power off. List price is \$4.95.



For further information, check 46U on Literature Card.

### Printed Circuit Soldering Kit

Ungar has introduced a new No. 270 "De-Soldering Kit" that includes several specially-shaped tips to simplify removal of components from printed wiring boards. One tip is bar-shaped for unsoldering several connections in a straight line; another is cup-shaped for unsoldering a tube socket in one operation; and still another is slotted for straightening component leads while they are being unsoldered.



For further information, check 47U on Literature Card.

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Morris Weisenberg, Chief Technician  
Universal TV Co., Los Angeles, Soys...

**"We Save 70% of Time and Temper  
In De-Soldering Printed Circuits..."**

"As one of the largest service organizations, with a fleet of 28 trucks, saving time for our 45 personnel is important. In the past, we found that regular soldering irons and tips, when used for de-soldering, would break off lugs or ruin the printed board.

Ungar's De-Soldering Kits permit the easy, quick removal of multi-lug components without disturbing printed wire continuity. Any part is speedily removed without damage."

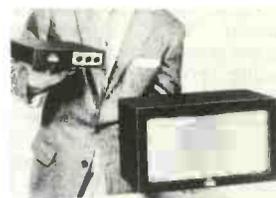
Write for complete details and new catalog

**Ungar Electric Tools, Inc.**

4101 Redwood Avenue, Los Angeles 66, California

### Stereo Conversion Kit

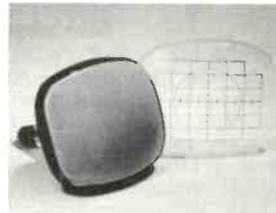
Walco has announced a kit for converting moderately-priced phonographs for stereo operation. Included are a 4-watt push-pull amplifier, two tone controls, a 6" speaker and cabinet with baffles, and a four-wire stereo cartridge. List price is \$59.95.



For further information, check 48U on Literature Card.

### 110° Check Tube

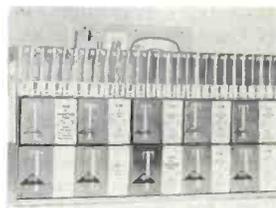
CBS-Hytron has introduced an 8" test picture tube, Type 8JP4, with 110° deflection and aluminized screen. The base is of the small-button (miniature-pin) type but can be converted to the shell type with an adapter. Other accessories are a patterned plastic mask and a carrying carton.



For further information, check 49U on Literature Card.

### Yoke Replacement Kit

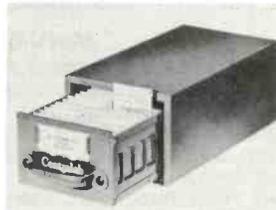
Triad is offering a "Yoke Pack" containing a replacement stock of most-often used 70° and 90° yokes and accessories. Included are 8 yokes, 25 network kits, yoke clamp CL-1, and centering cover YC-1. Space is provided at the top for storing the Triad 1958 instruction-book catalog.



For further information, check 50U on Literature Card.

### Ceramic Capacitor Kits

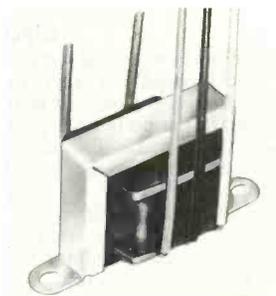
Centralab offers four new kits of "Hi-Kap" ceramic capacitors packed in steel file cabinets. Kit DCK-200 contains 200 tubular ceramic units, DDK-200 includes 200 standard disc types, TCK-80 contains 80 temperature-compensating capacitors, and HVK-150 includes 150 high-voltage discs.



For further information, check 51U on Literature Card.

### Transistor Transformers

Chicago Standard Transformer Corp. announces three new transistor transformers. TA-5 is designed to match a low-impedance microphone to the input circuit of a 2N156 or similar transistor; TA-16 is a driver transformer for class-B audio amplifiers employing a single 2N156 or a pair of 2N278's in push-pull; and TA-17 matches a class-B modulator consisting of a pair of push-pull 2N278's to a class-C load.



For further information, check 52U on Literature Card.

### Mylar Dielectric Capacitors

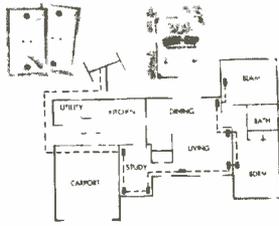
Aerovox has made available a new "Aerofilm" series of bypass capacitors (Type V84C) with Mylar dielectric and white ceramic cases. These units are claimed to have smaller size, lower power factor, and higher insulation resistance than previous types of molded-tubular bypass capacitors.



For further information, check 53U on Literature Card.

### Wall Receptacles

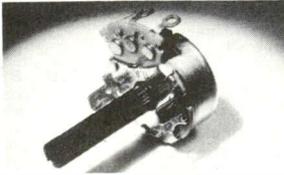
R-Columbia has introduced a "Switchmatic" wall receptacle for making antenna connections to TV sets. Several of these units can be interconnected and attached to one antenna in such a manner that the unused portion of the system is automatically disconnected when a TV set is plugged into any one of the outlets.



For further information, check 60U on Literature Card.

### Tab Mount Controls

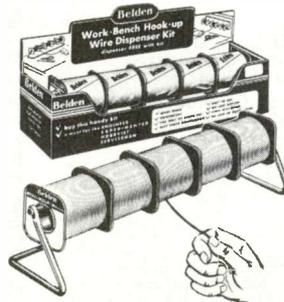
Clarostat is now supplying Series B47 tab-mount controls to the radio-TV replacement market. A selection of 25 different resistance values is available. Each control has a 1" phenolic shaft that is both knurled and slotted.



For further information, check 54U on Literature Card.

### Hook-Up Wire Kits

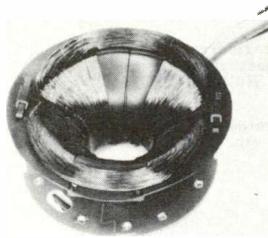
New hook-up wire dispenser kits are being offered by Belden. 14 different stock combinations are available; some kits have eight or ten 25-ft. spools of wire with different colored insulation, while various other kits have five 100-ft. spools. Each kit includes a dispenser rack that can be mounted either on the bench or on the wall. A choice of cellulose acetate, vinyl thermoplastic, or Teflon insulation is available.



For further information, check 55U on Literature Card.

### Replacement Yokes

Ram has announced availability of two new replacement yokes for 110° TV receivers. Type Y-110 is for Admiral 16F1 and related chassis, RCA Victor chassis KCS107 and -109, and Packard-Bell chassis 98D3. Type Y-111 is equipped with a damping network and plug for use with RCA Victor chassis KCS108, -113, -116, and -118. List price is \$22.50 each.



For further information, check 56U on Literature Card.

### Stereo Preamp

Grommes new Model 208 provides balanced outputs and necessary equalization through the use of gauged controls. When monaural material is being played, only one of the preamp channels is used to drive both power amplifiers in the system. Net price is \$99.75.

For further information, check 57U on Literature Card.

### Antenna Base

South River has announced a new "Bantam" TV tower weighing 3 lbs. and constructed of hot-dip galvanized tubular steel. This tripod-shaped unit is recommended for supporting an antenna on a 5' or 10' aluminum mast without guy wires. Thus loaded, the "Bantam" has withstood extensive wind resistance tests at velocities up to 90 mph.

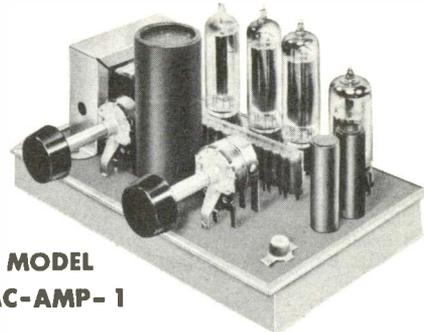
For further information, check 58U on Literature Card.

### P-N-P Transistors

Workman TV has announced a line of 19 p-n-p general-purpose audio and AF transistors, including power output and high-frequency types. The price list (Form No. CS-41) includes a complete cross reference between Workman transistors and comparable types bearing standard "2N" type numbers.

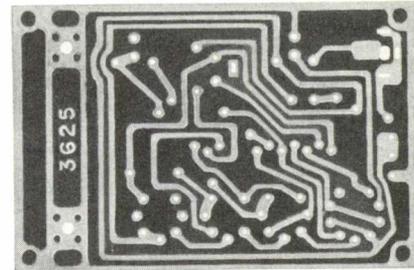
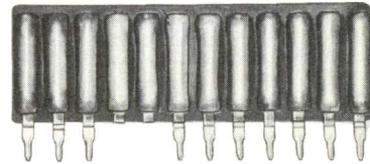
For further information, check 59U on Literature Card.

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### featuring "PAC" and an ERIE Printed Wiring Board

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- OUTPUT TRANSFORMER
- FILTER CAPACITOR
- VOLUME CONTROL and SWITCH
- TUBE SOCKETS
- CAPACITORS
- TONE CONTROL
- TUBES

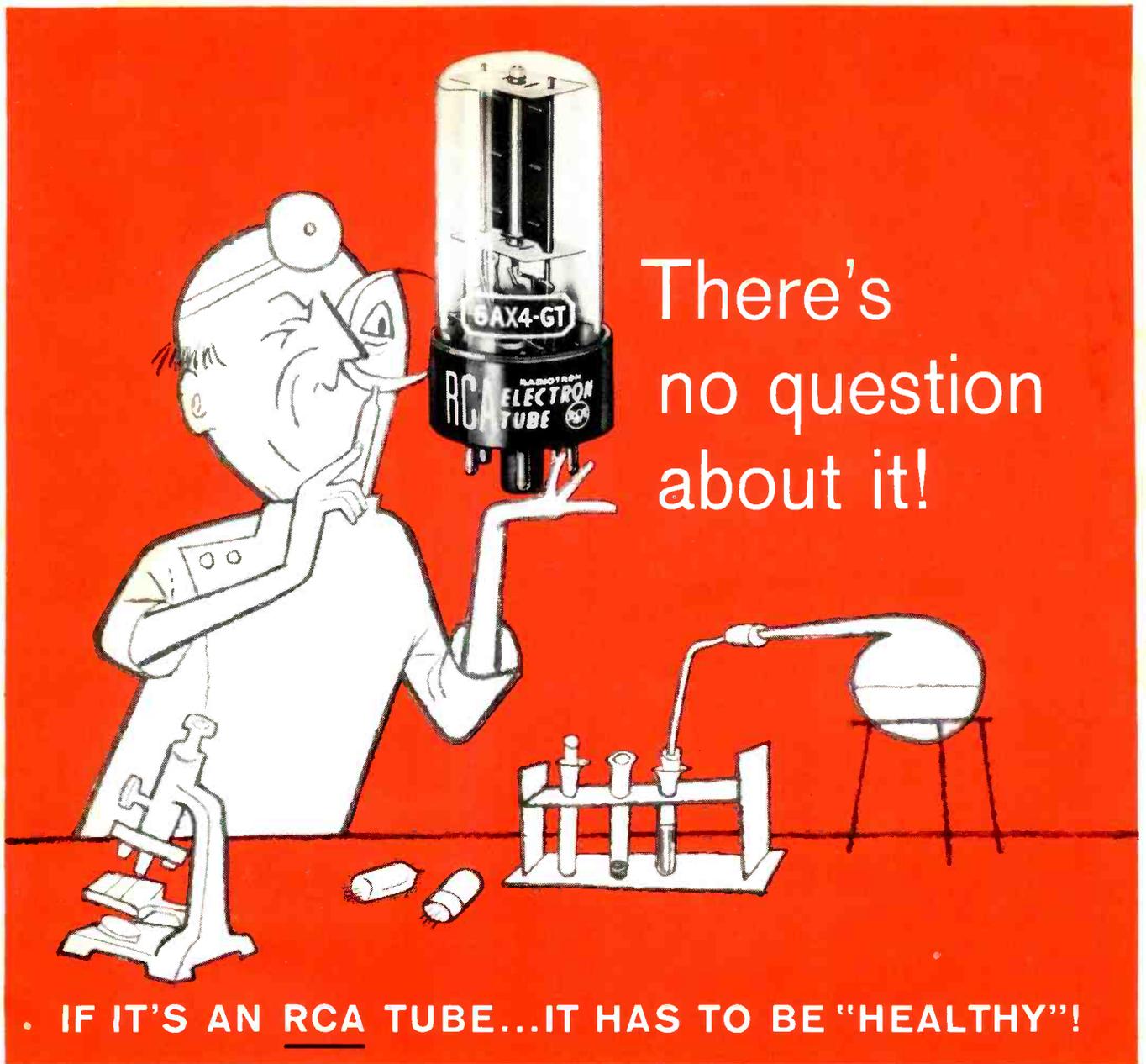
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- Frequency Response: 30 cycles to 12,000 cycles  $\pm 0$ , -3.5 db.
- Sensitivity: 0.56 volt RMS (input at 1 KC) for 2 watt output.
- Power Output: 2 watts • Input Impedance: 2 megohms.
- Output Impedance: 4 ohms • AC Power Consumption: 17 watts.
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**RCA Technical  
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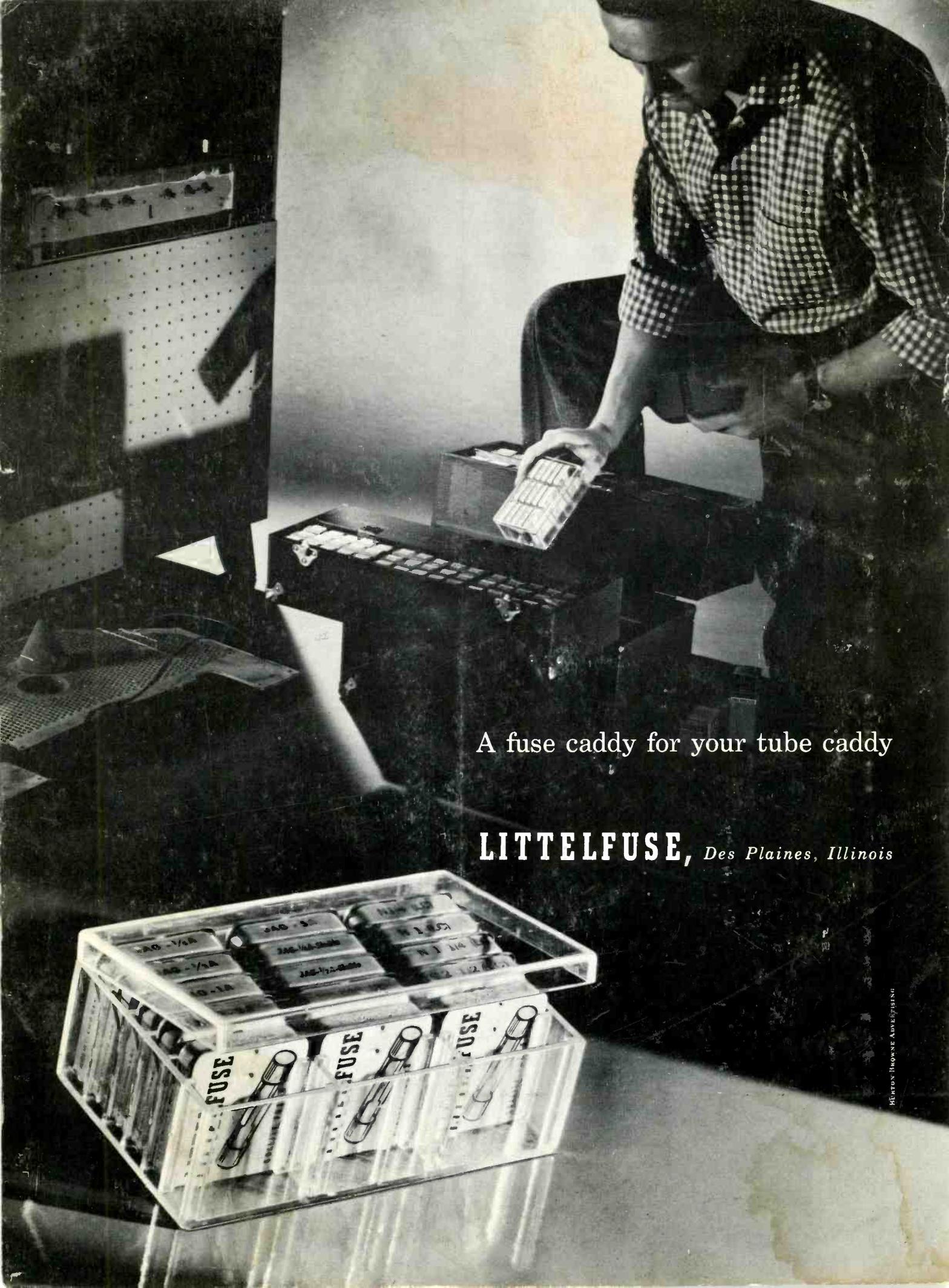
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