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Lead vs. hydrocarbon emissions

The manufacturers of electric vehicles and the batteries for those vehicles recently took a bashing from a report issued by Carnegie Mellon University. The report suggested that the contaminants released by the production and recycling of the necessary lead-acid batteries for powering the cars posed more of a threat to the environment than the emissions from gasoline powered vehicles.

But now the Electric Vehicle Association of the Americas (EVAA) has stuck back by refuting Carnegie Mellon's methodology and findings. EVAA's 13-page technical brief entitled "Environmental Impacts of Lead-Acid Batteries in Electric Vehicles" declares that the Carnegie Mellon study was seriously flawed and led to misleading conclusions.

EVAA points out that crucial calculations in the report were based on incorrect technical data that was derived from General Motor's electric-powered Impact car. That mistake set off an alarming series of miscalculations and unsubstantiated "guesstimating," adding "Those inaccuracies were compounded by the study's presentation of conflicting data in text and charts."

The EVAA report goes on to state that the Carnegie Mellon study grossly exaggerated the projections of the electric vehicle population just to exaggerate their environmental impact. It said that the projection of 10 million electric cars is nearly 80 times the number



THE GM IMPACT PROTOTYPE elctric vehicle

of electric cars (127,000) needed to meet the EPA zero-emission-vehicle mandates by the year 2000, which now apply in three states.

The brief asserts that the existing inventory of lead-acid batteries for the 188 million cars, trucks, and buses now on the road in the U.S. dwarfs all reasonable projections of increases that would be related to electric vehicles. It reasoned that even if each of those 127,000 mandated electric cars were powered by 27 lead-acid batteries (the complement of an Impact), those 3.4 million batteries would only represent less than 2% of all the lead-acid batteries now in cars in this country.

The EVAA refutation declares that the Carnegie Mellon report failed to take into account the probability that future electric cars will include more advanced batteries that contain less lead than today's products, and that materials other than lead can be substituted for electrodes. Moreover, the brief said no mention was made of other possible forms of electrical storage

for vehicles, such as fuel cells and flywheels.

To cap off its brief, EVAA zapped Carnegie Mellon for failing to recognize that most lead for car batteries is obtained from recycled lead products, and that modern lead recycling is essentially a closed-cycle, pollution-free process.

Smart cash cards proliferate

Many Europeans have become accustomed to paying for toll calls from public telephones with smart cards containing an embedded microprocessor that deducts value from the card as the caller speaks. The prepaid cards are purchased in local shops in various denominations, and they can be used as many times as it takes for the calling charges to equal the card's purchase price. This form of digital smart card has has now been upgraded Continued on page 10

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BY DAVID LACHENBRUCH

Recordable DVD

The battle over standards for the higher density CD-known in shorthand as the digital videodisc (DVD)-is intensifying, with both proponent groups calling for a single system, but neither willing to give an inch. As reported here (Electronics Now, August 1995), the systems, developed two bv Sony/Philips and Toshiba/Time Warner are mutually incompatible, and both are being urged not only as the next generation of video disc but as extremely high-capacity CD-ROMs for computers. The two systems are identified as Multimedia CD (MMCD) and Super Density (SD), respectively.

Struggling to compensate for SD's endorsements by movie companies and consumer-electronics manufacturers, MMCD has been pushing for backing by computer companies, but at our press time had received the endorsement only of Gateway 2000, the big directmarketing PC manufacturer. Now there is reliable information that the MMCD group will soon propose a recordable version of MMCD for introduction as early as 1997. The MMCD-R/E (for "record/erase") extension will use phase-change recording technology, with magneto-optical recording as an optional method. The SD group has already indicated that phase-change recordability is in its future.

Meanwhile, Thomson Consumer Electronics says that it plans to launch the video version of DVD at the May 1996 Consumer Electronics Show in Orlando under its RCA, GE, and ProScan brands. The company is forecasting sales of 2-million players in the first year of sales. Because of the decline in the value of yen, however, Thomson is now hedging its earlier prediction that they will retail at \$499 from the start.

Sony's "Micromirror" projector

In a three-year project, Sony has built a prototype high-definition projection-TV system using the Digital Micromirror Display (DMD) technology developed by Texas Instruments. Although neither company is talking about the subject, it seems likely that Sony eventually will produce a line of Micromirror projection sets for industrial, commercial, and consumer use.

TI calls DMD a "digital light switch." Each DMD has an array of hundreds of thousands of aluminum mirrors fabricated on a silicon chip 5/8-inch square. Each tiny mirror represents one pixel. When a voltage is applied to a mirror, it pivots slightly, reflecting light from an external light source through a projection lens onto a screen. When in the "off" state, light reflected from the mirror doesn't enter the lens. Projectors may use from one to three DMDs, with resolution as high as 1280×1024 pixels. The high-definition system built by Sony and was designed to represent the ultimate systemfrom which more simple systems could be devised.

Three other companies are preparing to produce non-consumer commercial versions of DMD projectors, to sell at \$5000 and up. TI, however, is hoping that a consumer version can sell at around \$2700, placing it at a popular projection-TV price.

The advantages of DMD, as summarized by TI, are as follows: DMDs can be fabricated on standard semiconductor production equipment. The system is all-digital and thus will be completely compatible with digital broadcasting, requiring no digital-to-analog conversion. It displays a complete frame at a time, without interlaced scan, thereby eliminating flicker. The picture is sharper and crisper, without blooming or color fringing characteristic of shadow mask tubes. The colors are mixed digitally for exact control and don't depend on phosphors. Brightness can be much greater than CRTtype projectors because an external light source is used. The system has square pixels for computer compatibility. Lower gain screens can be used, eliminating the directionality characteristic of CRT systems.

Consumer versions of DMD projectors could be available within two years, according to some manufacturers. The timing depends on the completion of accelerated life tests, which TI says have proven a five-year life span and should be able to guarantee the 10-year life span required by consumer products by this fall.

42-inch plasma TV

Last month we reported that Mitsubishi is planning to use plasma displays in its large-screen TVs as early as next year. Now Fujitsu is setting up a factory in Japan to *Continued on page 10*



OP-AMP ANTENNA Amplifier?

I have a long-wire antenna that I would like to share among several receivers. Given the wide bandwidth and high input impedance of FETinput-type op-amps (TL071/2/4 for example), could I use an op-amp as an cuit we breadboarded and tested.

In your situation, there are two problems to be solved: impedance matching and isolation. The impedance of a random-wire antenna at broadcast-band frequencies will be several kilohms, a poor match for the 50-ohm input of a receiver. Also, the inputs of multiple receivers can't be con-



FIG. 1—TL054 OR TL074 HIGH-FREQUENCY OP-AMP distributes broadcastband signals to several receivers.

impedance buffer for the broadcast band (535—1620 kHz) to fan out one antenna to feed the 50-ohm inputs of multiple receivers?—S. A. Jensen, Brooklyn, N.Y.

We were about to say "of course not" until we tried it! A few years ago, it would have been absurd to use an op-amp for RF signals—the frequency response would be inadequate and there would be too much noise. Today, however, there are good low-noise op-amps usable at frequencies of several megahertz. Figure 1 shows a cirnected directly to a single antenna, or they'll detune each other.

The circuit solves both problems. The first op-amp overcomes the impedance mismatch, strengthening the signal greatly even though it has no voltage gain. The rest of the op-amps feed the signals to the separate receivers. We used a Texas Instruments TL054 op-amp, but the TL074 would work just as well. (Not the TL084, though.) No low-pass filter is needed because the gain of these op-amps drops off sharply above 2 or 3 MHz. You may need to add some traps to cut out the signals of strong local stations as shown in the schematic. In fact, the overall noise level in your receiver will probably be lower after you add them, even if you weren't aware of a problem. You can make each trap out of any coil and capacitor that resonate at the desired frequency.

CUSTOMIZED REMOTE Control

I need a TV remote control for arthritic hands and fingers, one with larger buttons or rocker type volume and channel switches. Can you give me a diagram with the parts and values to construct one with just on-off, channel change, volume, and muting?—H. DeKeyser, Lake Elsinore, CA

Start by buying a universal remote control at a discount store. That will give you the ICs, LEDs, and other support circuitry readymade. Then mount it in your own enclosure and hard-wire your own switches for the functions you need, and leave out all the rest.

This is an approach that hobbyists should consider for lots of projects—instead of building something from scratch, modify a readily available consumer item. It's cheaper!

CACHE MYSTERY

I recently installed 256K of external cache RAM in my Leading Edge computer and now, although I can still





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Cache RAM is a small amount of special, fast memory into which the computer puts a copy of whatever it is reading from the real RAM. When executing a program, the CPU often reads the same small area of RAM repeatedly, and by fetching instructions from the cache instead of from RAM, it is able to run faster. Windows doesn't require any special setup to use a cache. In fact, software should not even be able to tell whether the cache exists.

The first thing you should do is replace the cache RAM chip; it's probably defective. If it checks out OK, and all your DOS applications run properly, the next step would be to start Windows with the command "win /b." That tells Windows to create a file called BOOTLOG.TXT describing every step it when through when starting up. That will at least tell you what part of Windows is running into the problem. You might be able to deinstall a device driver to clear up the problem.

TRS-80 FOREVER?

I have a Radio Shack TRS-80 Model III computer. It needs servicing, but I can't service it without the schematic diagrams. I asked Radio Shack and they replied, "We don't have those schematics, we threw them away a few years back." Where can I get them?—J. Reyes, Laredo, TX.

ast summer I bought a mostly working TRS 20 Model 4 for \$10. I need some technical information in order to use it effectively. Are any TRS-80 user groups still in existence?—S. Fuesting, 911 Keller, Effingbam, IL 62401.

Support line at 1-800-THE-SHACK and/or Tandy National Parts at 1-800-322-3690. Also, schematics and service manuals for TRS-80 computers are available (for about \$40 each) from Howard W. Sams & Co., 1-800-428-7267.

As for user groups, there's a thriving Tandy forum on the Internet in the newsgroup comp.sys.tandy. Also, we're publishing reader Fuesting's complete address so that TRS-80 Model 4 enthusiasts can contact him.

ELOATING A BATTERY

I would like to have a circuit that will maintain a 12-volt automotive battery fully charged without overcharging it.— J. C. Farley, Landover, MD.

called float-charging. To floatcharge a 12-volt lead-acid battery, all you need is an accurately regulated 13.8-volt power supply (such as Radio Shack's No. 22-120, or one of your own design).

You could connect the power supply directly to the battery, but it's safer to make the connection through a couple of 12-volt, 1-amp automotive



FIG. 2—LIGHT BULBS LIMIT CURRENT, float charging the battery without overloading the power supply.

light bulbs (see Fig. 2). The light bulbs serve as current limiters; they guarantee that no matter what happens, the power supply will never have to deliver more than 2 amps. Resistors can also serve as current limiters, but light bulbs are better because their resistance varies with current in a useful way. It's nearly zero at low current, but as the current through each bulb approaches 1 amp, the bulb lights up and the resistance increases. Thus, the bulb limits the current but doesn't waste energy when the current is low.

This setup is a good way to power a heavy load that operates intermittently, such as a ham radio transmitter. The battery delivers heavy current when necessary, and the power supply keeps the battery charged up.

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WHAT'S NEWS

continued from page 4

from simply prepaying phone calls to the rank of a full fledged electronic cash card.

The telephone digital smart card is inserted into a slot in special public telephone that deducts value from the card. Fumbling for the right change to make the call is eliminated as is the aggravation of searching for additional coins to "feed the box" if the call exceeds the minimum toll. The cards are especially convenient for children and visually handicapped persons.

Now both American and European companies are expanding that concept into what have been called electronic "purse" cards for making all kinds of payments. The card's current value is digitally stored on the card itself. Store clerks will use a smart-card terminal to deduct the amount of each purchase.

Last July two large British banks, National Westminster and Midland, began testing Mondex, a card that "stores" cash. Mondex cards are favored by merchants for making small purchases because they do not have to install a telephone link to a central computer. This card's drawback is that, unlike a credit card, losing the card is like losing cash.

MasterCard International, Visa International, and Bank of America are also developing electronic cash cards. MasterCard's smart card combines electronic cash with existing credit or debit cards. Visa will test its cards at the Summer Olympics in Atlanta next year. Electronic Payment Services expects to test an electronic cash card in Delaware in automated teller machines next July.

The Microsoft Corporation plans to enter the electronic cash field. It is working with semiconductor manufacturers to develop a card for use by banks that would underwrite the payments programmed into the card. The card is intended to replace cash with digital money that would be stored on the card and then be spent for small purchases not unlike the

prepaid meal tickets that were com-10 mon a half century ago.

Industry's fastest 1-Mb SRAM

A high-speed CMOS memory IC from IBM makes it possible to access its contents in 2.5 nanoseconds (ns) the time it takes a beam of light to travel a meter. This makes it the fastest 1-megabit synchronous static random-access memory (SRAM) in production. It is intended for PowerPCs and other high-performance RISC-based servers and workstations with speeds up to 200 MHz.

As central processing units (CPU) speed up, a workstation's memory must be able to keep pace in supplying it data. The advanced CMOS SRAMs permit workstation cache memories to match the speed of the CPU, significantly enhancing performance. Two 1-MB SRAMs can form a 256 Kbyte cache memory in a 200-MHz system.

The new IBM SRAMs provide enhanced data throughput in synchronous pipeline or flowthru operations, reduced signal noise with low-voltage TTL I/O level compatibility; and a self-timed late-write feature that allows consecutive read and write cycles. The 32 Kbit \times 36 and 64 Kbit \times 18 organizations minimize device count and optimize cache size.

ISCET receives White House recognition

President Clinton congratulated The International Society of Certified Electronics Technicians (ISCET) on its 25th anniversary. Founded in July of 1970, ISCET will celebrate its 25th anniversary at the National Professional Electronics Convention, July 31 to August 5, in Arlington, Virginia.

ISCET-certified technicians reside in all 50 states, and 41 territories and foreign countries. It is expected that the 40,000th technician will be certified before the end of 1995. The certification program has given consumers confidence that ISCET certified technicians have the knowledge, training, and experience needed to perform their service work reliably and professionally.

VIDEO NEWS

continued from page 6

make 42-inch thin plasma displays. The company plans to offer plasma TV sets in sizes from 20 to 100 inches diagonally starting in 1997, when Japan is scheduled to start commercial HDTV broadcasts, under its Fujitsu General brand in Japan. The company believes sets smaller than 20 inches will continue to use standard CRTs, while those bigger than 100 inches will be projection sets.

Fujitsu's factory will start producing 42-inch displays next year and plans to turn out 20,000 monthly to start. The company already sells 21inch displays for monitors, and has sold some plasma TV sets, but sales are limited by current high prices nearly 1,000,000 yen in Japan, the equivalent of almost \$12,000.

Two-program VCR

Hitachi has introduced a VCR in Japan that can record two programs simultaneously. The recorder is compatible with standard VHS tapes in the single-program mode. For twoprogram recording, the deck records both programs side by side in a standard 58-micron helical track, using a 25-micron head to record one program and a 19-micron head for the other. The recorder has eight video heads and two stereo audio heads. Two models have been introduced in Japan—one permitting simultaneous playback of both programs, either side-by-side or picture-in-picture.

Hitachi hasn't indicated whether it will offer the two-program recorder in the U.S., but it has introduced two unique video-recording products here. A Hi8 camcorder has a detachable shirt pocket-sized camera head that is attached by cable to the unit's body and can be used up to 21 feet away. The camera head is waterproof, so it can be lowered from a boat for underwater photography, for example, while the main part of the camcorder remains high and dry on deck. The camcorder also includes a 4-inch LCD monitor. EN

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LETTERS

SEND YOUR COMMENTS TO THE EDITORS OF ELECTRONICS NOW MAGAZINE

DANGEROUS METER SHUNTS

The use of No. 20 copper wire to make a 50-ampere meter shunt, as suggested in the May 1995 "Q&A," is an invitation to disaster. It is likely that the shunt would vanish in a flash of blue vapor and the meter would be destroyed.

Calculated values of fusing current are only approximations, because the actual value is affected by such factors as ambient temperature, air circulation, variations in wire diameter, and heat conduction through the connecting leads.

The calculated fusing currents for No. 20 and No. 22 copper wire are 58.4 and 41.2 amperes, respectively, so substituting the smaller wire would certainly prove disastrous. Even if it didn't self destruct, the shunt would run so hot that the meter readings would be unreliable. Moreover, soldered connections would probably melt.

The 50-ampere load device whose current is to be measured would have, if it conformed to code, leads of No. 10 wire or larger—perhaps No. 8 or No. 6. The shunt should be of comparable size. The one-foot length of No. 10 copper would have the resistance of 0.001 ohm required by the example.

The shunt should be located near the load, and sturdy mechanical binding posts, rather than solder joints, would be more suitable for the load connections. The meter leads should be connected inboard of the current leads, as shown in Fig. 3 on page 10 of the May issue.

Only the distance between the meter leads is effective in calculating the length of the shunt. The meter leads can be standard products of any convenient length. LEWIS F. GARBER Duarte, CA

With regard to the May 1995 "Q&A," the response to Steve Baumfalk's question was misleading. A No. 20 wire shunt for carrying 50 amperes would be a dangerous mistake; that wire would burn out in seconds! A No. 20 copper wire should not be trusted to carry more than about 5 amperes.

The National Electrical Code, written to promote safe installations in house wiring, requires at least a No. 6 wire—and that is for conductors in conduits. But this emphasizes the differences in current-carrying ability between No. 6 and No. 20 wire.

I would obtain resistance wire, ribbon, rod, or a standard shunt modified for that value. Fasten the calculated length into solid blocks or lugs with separate take-offs for the meter leads. Then trim the calibration with a series resistor, perhaps even an adjustable resistor, and calibrate it against an ammeter of known accuracy.

DWIGHT EGGLESTON Hendersonville, NC

STANDARD VALUE COMPO-NENTS ONLY

In the future, please do not specify non-standard values for components! I am referring to the Headphone Amplifier and Audio Chip described in the June 1995 "Q&A."

Forty years ago it was possible to find $0.04-\mu F$ or $0.05-\mu F$ capac-

5%	10%	20%
10 11	10	10
12 13	12	
15 16	15	15
18 20	18	
22 24	18	18
27 30	27	
33 36	33	33
39 43	39	47
47 51 56	47 56	47
62 68	68	68
75 82	82	08
91	02	
tors in a radio or television set but		

TABLE 1- RESISTOR VALUES

itors in a radio or television set, but components have become more standardized since then. A $0.047-\mu F$ capacitor should be specified in those applications.

It is always useful to remind authors and readers (especially those now learning the basics of electronics), that resistors, capacitors, and most fixed inductors are available only in standard values.

As a rule, there are six standard values per decade for components with 20% tolerances, 12 standard values per decade for 10% components, and 24 standard values per decade for 5% tolerances.

Capacitor manufacture offers one exception to this rule. Just because a capacitor is marked 5%, you should not conclude that there are 23 other capacitor values per

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GRANTHAM College of Engineering Grantham College Road Slidell, LA 70460 decade. Capacitor manufacturers confine themselves to six values per decade (and there are even exceptions to that rule deviation). This holds even if they offer a 10%, 5%, or even 2% tolerance.

Most capacitors are placed in RC circuits, and designers typically select the capacitor first before selecting a resistor that can be trimmed to the desired RC value. Thus, there is no demand for both capacitors and resistors in a full range of values.

Table 1 lists standard 5%, 10%, and 20% values. Remember that the table represents only one decade, meaning that 2.2-, 22-, and 220-ohm, 2.2-, 22-, and 220-kilohm, 2.2-megohm, etc. resistors are available. Potentiometers are typically available in only three values per decade: 10, 20, and 50.

You might be embarassed by asking for a 40-kilohm resistor at your local electronics retail store because you won't find one there.

DEAN HUSTER, INSTRUCTOR Francis Tuttle Advanced Technology Center

Oklahoma City, OK

VARIABLE DUTY-CYCLE TIMER

In the June 1995 "Q&A" column, Jan Janowski asked for a variable dutycycle timer with a fixed period of approximately 45 seconds. Q&A's answer implied that this was difficult for a single 555 IC timer.

The circuit shown in Fig. 1 gives an



FIG 1-CYCLE TIMER



FIG 2 - TIMER LOAD

output that is identical to the one shown in the "Q&A "column, yet only one 555 is needed. Moreover, there are no annoying duty-cycle "jumps" near the 100% *on* setting.

To give the appearance of 100% on to 100% off operation as Jan requested, simply insert an integrator between the timer and the load. Figure 2 shows an LED as the load with a capacitor acting as an integrator to give the illusion of 100% on to 100% off operation. The load could also be a solid-state relay (AC or DC) controlling a high-power device.

Control a lamp that way if you want a flashing light. By reducing the values of the timing and integrating capacitors (to eliminate flicker), you can make a fancy light dimmer! WILLIAM M. MILLER JR. North Dartmouth, MA

EASY DOES IT

Larry Klein deserves a lot of praise for his simple cable tester design that appeared the April 1995 "Audio Update." Even with 29 years of experience in electronic servicing, I am still looking for ways to make my work easier.

I maintain an audio-visual system on a U.S. military base in Germany. The air conditioning system in the auditorium is quite loud, so I have difficulty hearing the beeper on my digital multimeter when I am testing cables. The beeper is just not loud enough to be heard over the background noise. This made it impossible for me to test a long cable by myself.

Mr. Klein's cable tester equipped with a Sonalert brand alarm produces a sound that is loud enough to overcome the background noise. RICHARD J. MOLBY, WB7NZG/DA1DB HQ. USAREUR/7A, Germany SESCOM CT-6 CABLE TESTER KIT

EQUIPMENT REPORTS

This tester will verify that all of your cables are in good condition.

CIRCLE 15 ON FREE INFORMATION CARD



et's say that you have spent the better part of the day setting up audio equipment for a band that will be playing in a few hours and you have found that some part of the system is not functioning properly.

Mentally you run through a checkoff list of what could be wrong. High up on that list is the condition of the interconnecting cables. Could there be a fault in one of them? Do you have time to check out all the cables to identify the fault before show time?

Any intermittent cables—audio, video, or RF—can cause trouble, and it doesn't matter if the system is for a professional show or for a home entertainment system. A cable with an intermittent fault might behave well when you set up the equipment initially, with the fault developing some time later on during the performance or video playback.

Testing each cable before it is connected can eliminate problems once it's installed in a system. But checking the integrity of a lot of cables can be time-consuming unless you have specialized test equipment that will simplify the task.

Easy cable tester

The CT-6 cable tester from Sescom, Inc. (2100 Ward Drive, Henderson, NV 89015, 800-634-3457, 702-565-4828) is a compact matchbox sized instrument that will make short work of testing all different kinds of cable. The top cover of the CT-6 contains a matrix of pairs of standard jacks including 3.5 mm mini stereo, RCA, F, BNC, XLR, 1/4-inch stereo, and five-pin MIDI.

To test a cable, you plug each end of it into a matching jack on the tester and, depending on the configuration of the cable, up to three LEDs will light to indicate that there is continuity in the cable. Cables and their connectors should be wiggled and flexed while testing them to screen out any that might be making connections only when not subject to mechanical or thermal stress.

With the exception of the XLR jacks, where one is male and the other is female, all other jacks on the CT-6 are female. Thus any cables terminated with jacks will require an adapter. Unfortunately, if you want to test cables whose plugs or jacks don't match any of the connectors on the CT-6, you must purchase or make adapters for them. However, with satisfactory adapters, you can test unusual or uncommon cables with up to three conductors just as quickly and easily as standard cables.

The CT-6 tests cables with up to three conductors. (There are only three conducters in a five-pin MIDI cable.) The tester has a row of three momentary pushbutton switches on its top panel with an indicator LED built into each switch. The switches break connections between the two ends of a cable when pressed. The LED above each switch should turn off when the switch is pressed, indicating the discontinuity. If an LED doesn't turn off, this is an indication that there is a short circuit in the cable under test

All of the jacks and plug are soldered directly to one of two PC boards within the CT-6. This makes building the tester easier and less time consuming. It also eliminates a lot of internal wiring that could cause of problems by themselves and this form of construction makes the CT-6 an extremely rugged unit for use in the field as well as on the lab bench.

The CT-6 is packaged in a rugged, modern, black-anodized aluminum case that measures only $5^{1}_{2} \times 3^{3}_{4} \times 2^{1}_{2}$ inches. It is tough enough to stand up to a lot of abuse. The top panel is clearly labeled with silk-screened white lettering. The side panels of the case extend above the height of the top surfaces of the jacks to protect them in the event that a flat heavy object is dropped on it or if the instrument should accidentally fall to the floor. Four rubber feet on the bottom of the case prevent the unit from sliding around on the benchtop when it is being used.

Build it from a kit

A really attractive feature of the CT-6 cable tester is its low price of

only \$50 for a kit that includes the aluminum case. But building the kit should be fun and the results will be satisfactory even for relatively inexperienced hobbyists if they follow the directions carefully. The kit includes all of the components, circuit board, solder, and even a pair of fresh batteries. However if you don't have the time or would rather just purchase the cable tester complete, a fully assembled and tested CT-6 is priced slightly higher at \$89.

To build the CT-6, you install the jacks and switches on two small PC boards and solder three small preformed jumper wires in place. The two boards are interconnected with a flexible multiconductor cable, which offers higher reliability than if individual wires were used to make the connections.

The PC boards mount securely to the front panel. A battery holder snaps into the back panel of the cable tester and two wires are solderd between it and one of the PC boards on the panel. The top and bottom panels then slide into a partially assembled case. Four slotted corner rails, which mount to the side panels with screws, hold the top, bottom, and side panels in place.

Our CT-6 should have worked immediately after we finished it, but it didn't. After investigation we noticed that an unused screw-on steel collar from one of the jacks was lodged inside another jack. It was only after checking and rechecking the boards for the usual solder shorts, and then peering inside each jack that the unwanted interference was noticed and removed. The moral of this story is that little details can sometimes trip up even the most experienced electronics hobbyists.

The CT-6 worked fine after that. We can't blame anyone but ourselves for not noticing the problem sooner. Even if the construction manual had included one of those boring boilerplate paragraphs "carefully inspect all the components before doing any assembly" we (like many other hobbyists) probably would have ignored it. But don't let our experience put you off building the CT-6!

How to use the CT-6

Pop the two AA cells into the holder in the back panel and it's ready for use. The instrument has no power switch. To test a cable, simply plug the cable into the appropriate jacks and that turns the CT-6 on. The number of LED lamps that will light depends on the construction of the cable under test, typically two or three conductors. The pushbutton below each LED on the panel should turn the LED off; if it doesn't turn off, suspect a short circuit. Flex the cable throughout its length to see if that has any effect on the integrity of the connection.

It is our view that those who regularly install cables in their daily work can avoid having to track down faulty ones by testing them first with the CT-6. It is clearly a useful addition to your bench tools or field repair kit. Considering the cabe testers reasonable price, exceptional versatility, and extreme durability, we strongly recommend the CT-6 as a test tool that will come in handy for many years to come.

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monitors include 0.27-mm dot pitch, 1280×1024 resolution with up to 69-kHz horizontal scan frequency, and a flat, square CRT screen. A wide tilt-swivel angle stand is supplied.

A dome speaker system in the monitor produces sound that is directed to the front of the monitor through acoustic ducts and projected through micro-hole speaker grilles. Because the sound is projected very close to the CRT, sound localization is enhanced.

The monitor's icon-based, onscreen display menu permits picprinted page. Proprietary coating reduces static electricity, screen glare, and reflections from the monitor for easier viewing

The speakers project only $1_{/4^-}$ inch on each side of the monitor. The high-sensitivity, unidirectional microphone, located in the center of the lower chassis, can voice-activate software and initiate video teleconferencing and audiorecording devices. Two jacks are provided for optional headphones and an additional microphone.

An auto-switching universal power supply permits the monitors

to be powered from a 50 to 60-Hz source of 90 to 264-volt AC. The monitors conform to VGA, SVGA, XGA, Mac, and VESA standards, as well as the EPA's Energy Star guidelines. The monitors are certified for MPR II low magnetic radiation.

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Continued on page 29

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All fieldpacks include a 300-ampere AC current clamp, a dual-temperature accessory with K-type thermocouples, test leads with removable tips, an alligator clip, and an extra set of long probe tips. The accessory heads measure temperature, AC current, DC microamperes, and relative humidity

When carried by its shoulder strap, access to any item in any pocket can be gained from the top. A zippered outside pocket can hold paperwork or a notebook computer. The inside pockets have clear plastic faces permitting their contents to be seen at a glance.

The HS24K14 fieldpack includes a heavy-duty HS24 "stick" meter and the DC microampere head. The HS24K15 adds the relative humidity head and the HB73K13 includes the heavy duty HB73 DMM. The HB73K14 includes the relative-humidity head.

The HS24 DMM has the 12 most popular measurement ranges for field use. The HB73 DMM adds 18 switch positions and measures current up to 20 amperes with a resolution down to 0.1 microampere.

HS24K14 and HB73K13 fieldpacks are priced at \$319.00, while the HS24K14 and HB73K14 fieldpacks are priced at \$418.00 **FIELDPIECE INSTRUMENTS, INC.**

231 East Imperial Highway, Suite 250 Fullerton, CA 92635 Phone: 714-992-1239 Fax: 714-992-6541

STUDENT OPTICAL BENCH.

THE ES STUDENT OPTICAL Bench from Edmund Scientific is a scientific teaching aid for demonstrating all the basic principles of light propagation. Experiments performed with the bench give students hands-on experience in building such basic optical instruments as telescopes, microscopes, and magnifiers in the classroom or at home.

The basic bench includes a multi T-slot meter beam with a stabilizing support and a full 1000-mm scale; a lamphouse with a bright filament bulb, socket, and DIN connector. In addition, it contains a 12-volt, 1.3ampere wall outlet AC to DC adapter; two magnetic stainless-steel carriers to accommodate a 50-mm-diameter lens and/or a 50-mm square slide. Also included are a 6-inch high, 12-inch wide plastic screen with combined X-Y-mm scales and 5-mm grid pattern.



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The kit, packaged in a compact case, also contains an assortment of 50-mm lenses and mirrors, 50-mm square slides (including filters), diffraction gratings, polarizers, slotted and round aperture plates, a diffuser, and an object slide. The instruction manual included describes nine experiments that introduce students to the concepts of geometrical optics, optical instruments, the polarization of light, and diffraction gratings.

The ES Student Optical Bench (Model 39,510) is priced at \$250.00 EDMUND SCIENTIFIC COMPANY Department 15A1 C999 Edscorp Building Barrington, NJ 08007-1380 Phone: 609-573-6886 Fax: 609-573-6295

DPM WITH LCD DISPLAY.

THE DMS-30LCD-4/20S 3-1/2digit, digital panel meter from Datel is loop powered. It includes an eightposition DIP switch for selecting the DPM's input range, zero offset, and decimal-point location. No jumpers are required.

The DMS-30LCD-4/205 measures



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September 1995, Electronics Now 2

www.americanradiohistory.com



 2.2×0.93 inches and it has a behind the pane depth of less than 1 inch. Its LCD display has 0.4-inch high char-



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acters. The fully self-contained DPM is powered by a 4 to 20-milliampere current loop, so it requires no external power supply, bypass capacitors, or trimming potentiometers.

DMS-30LCD-4/205 DPM is priced at \$74.00 in single units. **DATEL, INC.** 11 Cabot Blvd. Mansfield, MA 02048 Phone: 508-339-3000 Fax: 508-339-6356

FILER/SANDER.

THERE ARE TIMES WHEN electronic hobbyists and professionals must do some filing and sanding to build or repair circuitry or enclosures. Wahl's AC-powered Filer/Sander is a reciprocating tool for sanding and filing that relieves the user of the wrist stress during the repetitive motions of sanding and filing.



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The filer/sander is suitable for working on material as soft as balsa wood, and as hard as metal, as well as all kinds of plastics. It provides easy

access to hard-to-reach location on

the projects. It is sold with 19 accessories in a combined storage and carrying case.

The Wahl filer/sander is priced at \$78.95.

WAHL CLIPPER CORPORATION P. O. Box 578 Sterling, IL 61081 Phone: 1-800-735-9245

PEN-STYLE DIGITAL MULTIMETER.

WAVETEK'S MODEL DM73A pen-style digital multimeter is designed for single-handed use. This allows the user to concentrate on making measurements, especially in restricted spaces. It has a 4200-count display that provides users with exceptional accuracy and resolution for set-



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ting voltage levels.

The meter offers auto reading hold or probe hold which automatically freezes the reading on the display for later evaluation. The probe-hold feature is suitable for high-voltage troubleshooting and making readings of PC-board voltages. The reading record feature records maximum voltage and minimum resistance, permitting the tracking of variable voltage sources and general troubleshooting. The meter has a plug-in test lead with probe attached.

The Model DM73A pen-style DMM is priced at \$69.95, and an optional leather carrying case (Model VC13A) with snap closures and belt loop is priced at \$17.95.

WAVETEK CORPORATION

9145 Balboa Avenue San Diego, CA 92123 Phone: 619-279-2200 Fax: 619-540-0325

WRIST WATERBED.

WAVECO'S WATERBED FOR Your Wrists is a bar shaped pillow filled with a high-viscosity fluid that conforms to the user's wrists while he or she is engaged in long-duration data entry on a computer keyboard. It is located on the desk or bench in



CIRCLE 26 ON FREE INFORMATION CARD

front of the keyboard where it will ease stress on the operator's wrists by supporting them properly.

The Waterbed For Your Wrists reduces or relieves discomforts likely to occur with prolonged computer keyboard operation. Its gel-filled cushion makes a comfortable wrist rest and the wrists are supported even while entering data. It is to be distinguished from conventional wrist rests that function only as wrist support platforms during periods of inactivity.

The gel conforms to the contour of the wrist and palms to maintain close contact regardless of the position of the user's hands and arms. That feature prevents the wrists from contacting any hard corners or edges on the desk or keyboard which can result in *Continued on page 130*

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Low-Cost Software For Ecceptions

New low-cost software lets you draw schematics, lay out PC boards, and breadboard circuits on your desktop PC.

THERE WAS A TIME. NOT too long ago, when making a printed circuit board required a drafting table. custom drawing templates. decals, hobby knife, plenty of Band-Aids, and a darkroom. And if you wanted to test and debug the circuit, there was the extra step of breadboarding. Thankfully, that era is all but a faded memory. Today you can draw a schematic, breadboard the circuit, and design a foil pattern for a PC board without ever lifting a pencil or getting your hands dirty. You can do it all with software that you can run on your personal computer.

Circuit design software packages have been around for about 20 years. However, until recently, they were typically high-priced industrial applications that were out of the budget range of most hobbyists and many small businesses. The last couple years have seen an influx of low-cost circuit-design software that's ideally suited for home use and small-scale production work.

Because of the wide range of features available. finding the program that's right for you can be more of a chore than constructing a circuit board the old fashioned way! To take the risk and mystery out of shopping for these gems, *Electronics Now* scoured the market in search of circuit-design packages for schematic diagramming, PC

layout, and ordik circuit simulation costing \$350 and less. Then we tested them to see how they compared in price. features, and ease of use. Before we get to the software-rendered artwork. take a look at Fig. 1. Shown is a diagram of the Electronics Now benchmark circuit as drawn by the Electronics Now art department. It can be used to compare the quality and aesthetics of the schematics drawn by the circuit design pro-

board

grams reviewed here. The circuit, which is a functional sawtooth generator, was selected because it's simple enough to draw and construct, yet complex enough to be repre-

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TJ BYERSit Circuit Wind

- Basic D

cuit Board

ulation...





FIG. 1—THIS CIRCUIT WAS DRAWN by the *Electronics Now* art department. Compare it with the quality and aesthetics of the schematics drawn by the circuit design programs reviewed here.

sentative of the type of projects you're likely to tackle. For example, it consists of three ICs, a popular small project count, that are a mixture of analog and digital devices. While the parts are commonly available, they are diverse enough so that nearly half of the software packages didn't have all three "in stock" in their parts libraries. That provided an opportunity for testing their component library editors. When available, text in the form of descriptive headings (clock, waveform generator, and low-pass filter) was added to the drawing.

Schematic software

This article looks at 13 lowcost programs that let you draw schematics with your PC. The schematic is displayed on the screen using standard symbols stored in a component library. To draw a schematic, you select the desired components from a library menu or dialog box, then position them on the screen with a mouse or keyboard. Ideally, the library should contain all the devices you'll ever need. In practice, expect to create new parts on an as-needed basis with the component editor. Devices are wired together by drawing lines between the appropriate pins or their connecting intersections.

Changes to the schematic are made with a screen editor, which lets you add, delete, and move both wires and components. Editing tools can make those changes easier. The better only. The remaining seven are schematic capture programs with netlist output for interface with PC board layout and/or circuit simulation software. The programs are packaged in three different ways. A few,

produce netlists-a list of de-

vices and their connections

with each other. Six of the pro-

grams detailed here are draw-

especially the schematic-draw programs, are stand-alone packages, which may or may not generate netlists. In some cases, the schematic draw software is bundled with companion PC board layout software,



FIG. 2—DESPITE ITS LOW ASKING PRICE OF \$20, Schematic Plus is quite flexible and easy to use. However, the library is limited to about a dozen components, so you'll have to create most of the devices yourself.

programs have a "rubberband" feature that lets the wires stretch and shrink as parts are moved on the screen, thereby maintaining their electrical connections. Without rubberbanding, only the device itself moves, forcing you to erase the old wires and redraw them. Other labor-saving features include auto annotation, which automatically assigns part references (R1, R2, etc.), and onthe-fly zoom or auto scrolling (called pan) for accurate placement of symbols on drawings that are larger than the screen.

There are two types of schematic diagram programs, those that just draw and those that which has the look and feel of the schematic draw program, but has no link to the schematic. At the top of the line are integrated schematic capture and PC board layout programs that work interactively. The better schematic capture programs, both stand-alone and integrated, have design-verification features that check the circuit for short circuits, opens, and invalid connections (such as the output of a TTL gate tied to ground).

Schematic drawing

Schematic draw programs differ from schematic capture Continued on page 137

YOU MIGHT BE SITTING ON A WINDfall of free and easily reusable semiconductor memory. For example, old motherboards, memory boards, and computer peripheral devices contain perfectly serviceable memory just waiting to be recovered. One problem is that today's personal computers no longer use DIPpackaged memory. If only the discarded devices could be recycled and put to use in the more modern single-in-line memory modules (ŠIMMs) required by the latest computers! As a matter of fact, they can. and this article will tell you how.

By following the techniques described, you can recycle old DIP-style DRAMs into 30-pin SIMM modules, at a cost of under \$10 per megabyte (excluding the cost of memory). With new 1-megabyte SIMMs selling for around \$40 apiece, the economics of recycling memory are obvious.

This article provides construction details on three DIPto-SIMM conversion boards that can be used in four different configurations:

1. 256 kilobyte (KB or K) consisting of nine $256K \times 1$ DRAMs.

2. 256 kilobyte consisting of two 256K \times 4 DRAMs and one 256K \times 1 DRAM.

3. 1 megabyte (MB) consisting of nine MB \times 1 DRAMs.

4. 1 megabyte consisting of two 1MB × 4 DRAMs and one 1MB × 1 DRAM.

The article provides background information on memory in general, construction details, and PC board patterns. Kits are available from the source listed in the Parts List. Figure 1 shows several standard SIMM modules along with the DIP/SIMM conversion modules for comparison.

SIMM city

A SIMM is a module that houses multiple DRAM ICs organized into bus widths that are convenient for specific computer architectures. Standard manufactured SIMMs are available in 30- and 72-pin configurations. The 30pin modules typically have databus widths of nine bits; the 72pin modules have data-bus widths of 36 bits. The conversion boards described here have the popular 30-pin format.

Actually, 30-pin SIMMs are available in both eight- and nine-bit data-bus widths. Ninebit modules are installed in most PCs, where the extra bit performs parity detection. To save money, some PCs (for example, some Compaq models) do not have the parity bit so they use 8-bit modules. Macintosh computers typically use the eight-bit format as well. The conversion boards described here will operate in either the eight- or nine-bit mode. To operate in the eight-bit mode, the parity checking IC is simply left out.

Generallly, 30-pin SIMMs hold a maximum of 16 megabytes; there are not enough address lines for more memory capacity. The boards discussed here can hold either 256K or 1MEG of memory.

Thirty-pin SIMMs are 3.5 inches long and less than an inch tall. To accommodate the larger ICs, the boards discussed here are slightly taller. The nine-IC boards are 1.5 inches tall, and the three-IC boards are 0.8

RECYCLING MEMORY FOR YOUR PC

Save bucks by recycling DIP memories into SIMMs you can make yourself.

LEON W. SCHMIDT



FIG. 1—DIP TO SIMM CONVERSION BOARDS are slightly taller than factory-made SIMM-only counterparts.



COMPONENT SIDE of the 256-kilobyte/1-megabyte board.



SOLDER SIDE of the 256-kilobyte/1-megabyte board.

inch tall. Before committing to this project, be sure that the host motherboard can accept the larger modules. You'll find it easier to work with motherboards that have their SIMM sockets mounted directly behind the keyboard connector.

A SIMM connects all address lines to all ICs in parallel. In addition, a SIMM parallels the control signals: row-address strobe (\overline{RAS}), column-address strobe (\overline{CAS}), and write enable (\overline{WE}). The data lines (DO-D7) come out separately, as does the parity data bit and its \overline{CAS} .

Figure 2 is the schematic of the 256K/1MEG combination board, Fig. 3 is the schematic of the 1MEG nine-chip board, and Fig. 4 is the schematic of the 256K, nine-chip combination board.

A DRAM stores a bit of data as a charge on a tiny capacitor. Each DRAM contains an array of storage capacitors. Access to any particular bit in the array is determined by the row and column indexes of that bit. Those indexes are set by the address lines. In addition, the state of the \overline{wE} line determines whether data is written to or read from a DRAM: a CPU reads data from a DRAM when \overline{WE} is high, and writes to it when \overline{WE} is low. DRAMs also have separate input (DI) and output (DO) pins.

Over time, the charge stored Continued on page 149



CARBON MONOXIDE DETECTOR

An early warning of the presence of carbon monoxide can save your life.

opaque surface is defined by the ratio of the light reflected from the surface to the total incident light, and the value is always less than unity.

The reflection coefficient can refer to a specular reflection or a diffuse reflection (see Fig. 1-a). In general, the coefficient varies with the angle of incidence, with the wavelength of the light, and the surface material. In the circuit presented here, the angle of incidence is held constant and the wavelength of the light (in the infrared range) is constant; the only variable is the surface.

A cross section of the surface of the CO reagent is shown in Fig. 1-b. Incident light enters the reagent particle film. Some of the light is reflected specularly and some is reflected diffusely, scattered by the reagent particles. If the reagent darkens due to the presence of carbon monoxide, the light reflected from it will decrease. The circuit senses this reduction in reflected light.

The CO detector uses an inexpensive chemical carbon monoxide detector card that is sold under the trade names "GAS ALERT plus" and "heads up!" as the sensor element. These sensors darken to a gray or black color when exposed to CO. The cards are a little over two inches square and have a light yellow colored reagent located in their centers (see Fig. 2). This project uses an optical sensor to con-

ROBERT GAFFIGAN Jr.

CARBON MONOXIDE. AN ODORLESS and colorless gas, is a naturally occurring byproduct of combustion. Carbon monoxide, or CO, is inert except for its strong tendency to combine with hemoglobin in the blood. Hemoglobin's affinity for carbon monoxide is 300 times that of oxygen. As a result, lethal levels of the gas in blood can be reached within minutes of exposure. The data in Table 1 make it very clear that an early warning of CO concentrations in the air can save lives.

This article describes a portable CO detector that can be carried to the job site or used in a tent, camper, boat, motel room, or anywhere else CO buildup could be a danger. The idea for this article came from a 1966 patent. The patent has since ex-

pired, but the method for detecting CO is still viable using 90's technology. The patent was based on a chemical-type CO detector that changed color with exposure to the gas. The detector described in the patent sensed a transmittance change through a CO-sensitive reactant. The 1960's technology depended on a photodetector to detect the transmittance change, but the photodetector's output voltage drifted with temperature. Moreover, the device was power hungry, which precluded battery operation.

The scheme described in this article to detect CO is based on a changing reflection coefficient of the reagent, rather than a transmittance change through the reagent. The reflection coefficient or reflectivity of an

Percent	Parts per million	Results
0.02	200 PPM	Slight headache within 2-3 hours
0.04	400 PPM	Frontal headache within 1-2 hours
0.08	800 PPM	Dizziness, nausea and convulsions within 45 minutes. Insensible within 2 hours.
0.16	1600 PPM	Headache, dizziness and nausea within 20 minutes, death within 2 hours.
0.32	3200 PPM	Headache, dizziness and nausea within 5 to 10 minutes, death within 30 min.
0.65	6400 PPM	Headache, dizziness in 1 - 2 minutes, death in 10 - 15 minutes.
1.28	12,800 PPM	Death in 1-3 minutes.

TABLE 1 CONCENTRATION OF CARBON MONOXIDE

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FIG. 1—A REFLECTION COEFFICIENT can refer to a specular reflection or a diffuse reflection (a). The coefficient varies with the angle of incidence, with the wavelength of the light, and the surface material. A cross section of the surface of the CO reagent is shown in b.



FIG. 2—THE CO DETECTOR uses an inexpensive chemical carbon monoxide detector card. The sensor darkens to a gray or black color when exposed to carbon monoxide.

stantly monitor the card for discoloration, and it sounds an alarm if any darkening occurs.

Circuit operation

This circuit, shown in Fig. 3, has a focused optical sensor to transmit light to the CO re-



FIG. 3—CO DETECTOR SCHEMATIC. The circuit uses a focused optical sensor to transmit light to the CO reagent, and then it senses the amount of light reflected.

agent, and then it senses the amount of light reflected. The reagent for CO detection darkens from a light yellow color when exposed to CO. According to the card's manufacturer, a concentration as low as 100 parts per million (ppm) will darken the detector after 15 to 45 minutes. A concentration of 600 ppm will darken it in one to two minutes. The reagent will return to its original color when the air freshens, usually after about 10 minutes. The time it takes for the reagent to return to its original color depends on the concentration of CO to which it was exposed.

The "brain" of the CO detector circuit is a PIC16C71 microcontroller (IC2) that contains a built-in four-channel analog-todigital converter. Other than the microcontroller, the main component of this circuit is MOD1, an Omron EE-SY148 optical reflector module. The module directs infrared light to the CO reagent and then receives the light reflected back to it. When CO is present, the reagent will darken, thus reducing the reflectivity. The reduction is sensed by the microcontroller which then turns on buzzer BZ1.

Ideally, the reflectivity of the CO reagent should not change in fresh air, but because of the scheme used to bias the optical reflector (resistors R1-R4 and Q1), the reflector becomes very sensitive to temperature variations. Moreover, each detector reagent card has different shaped particles that are orientated differently. That causes the light hitting the reagent to scatter differently so that each card can have a slightly different reflectivity. The microcontroller has an automatic calibration routine that takes reflectivity and temperature Continued on page 153

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OVER/UNDER VOLTAGE PROTECTOR

This simple protective device can prevent costly damage to appliances that are always turned on.

RON TIPTON

THIS PROJECT WAS DESIGNED IN REsponse to a "power line inci-dent" in which a car or truck knocked down a telephone pole and short circuited a couple of the power lines. The line voltage rose to more than 150 volts over a significant part of the service area for at least 30 minutes. The next morning, many fax machines and PCs that had been left on were found "dead." This article describes a device that provides transient as well as under- and over-voltage protection for a fraction of the cost of an uninterruptable power supply (UPS).

The heart of the unit is a normally open solid-state relay that is held closed by the output of a window-type voltage comparator. If the comparator's input voltage goes outside the window for even a brief period, the relay opens and disconnects the load. Power stays disconnected until a reset button is pushed. If the line voltage is still too low or too high when the reset button is pushed, the load will again be disconnected immediately. A brief power outage of a tenth second or less will not disconnect power.

How it works

A schematic of the circuit is shown in Fig. 1. The ICL7665



Over/Under AC Line PROTECTOR

RESET

PUSH TO

TEST

INPUT

TDL

OUTPUT

FIG. 1—SCHEMATIC OF THE PROTECTOR CIRCUIT. The ICL7665 over/under voltage detector monitors the line voltage and disconnects the load if the input voltage goes below 95 volts or above 130 volts.



FIG. 2—PARTS-PLACEMENT DIAGRAM. Only one jumper wire is required in the position marked "J." The PC board is designed to accept a series pair of resistors for R2, R5, and R7 to make it easier to find a pair of 1% resistors that will combine to make the correct value.



over/under voltage detector manufactured by Maxim (IC1) monitors the line voltage and disconnects the load if the input voltage goes below 95 volts or above 130 volts. The ICL7665 incorporates a pair of comparators and a voltage reference into a single 8-pin DIP. The IC's switch points are accurately set when the chip is manufactured, and its outputs will drive TTL loads directly.

The over/under voltage detector (IC1) monitors the filtered, unregulated DC voltage developed across C5 which changes with the input line voltage. The input voltage levels to IC1 are set by resistors R2 through R8. Both comparators in IC1 change state at 1.3 volts, but the under-voltage output has a built-in inverter which provides a window comparator. As long as the voltage at pin 3 is above 1.3 volts and the voltage at pin 6 is lower than 1.3 volts, both outputs (pins 1 and 7) are low and RY1 is turned on and its contacts are closed. The resistor values for R2 through R8 were carefully chosen to set the voltage levels at which the circuit turns on and off. The math used to determine the resistor values is explained in the sidebar "Calculating Resistor Values." Onepercent metal-film resistors must be used for this network. Average line voltage varies from different locations. Trimmer potentiometer R4 provides a small adjustment range so you can set the window's center for your particular line voltage.

When IC1 detects an overvoltage at pin 3, it immediately opens the relay by sending a high output to pin 2 of NOR gate IC3-a which, in turn, turns off Q1. When IC1 detects an overvoltage at pin 6, the output at pin 7 goes high. That high does not turn the solid-state relay off directly. If it did, the energy stored in capacitor C5 would decrease slightly, which would make the circuit "think" that the input voltage has gone back within the desired window. Then RY1 would turn back on even though a voltage outside the trip point might be present. Continued on page 143 The bridge circuit holds an honored place in the history of electronic instrumentation. Learn how and why they make precise measurements.



A BRIDGE CIRCUIT IS AN INSTRUment for making comparison measurements. These circuits are widely used to measure resistance, inductance, capacitance, and impedance. They have four sections or arms connected in series to form a diamond. An AC or DC voltage source is connected between one pair of opposite junctions, and an indicating meter or output circuit is connected between the other pair of opposite junctions.

Bridge circuits operate on a null-indication principle. When the bridge is balanced, the output is zero. Consequently, the indication is independent of the calibration of the indicating device. For this reason, accurate measurments can be made with bridges. Bridge circuits can also control other circuits. When functioning as controls, one arm of the bridge contains a resistive element that is sensitive to such physical variables as temperature or pressure.

Bridge circuits are classical.

RAY MARSTON

electrical measuring instruments predating electronics as a field distinguished from classical electrical engineering. Until about 25 years ago, bridge circuits were the only practical instruments for making many kinds of precise measurements.

However, in many applications they have been displaced by digital multimeters that contain provisions for making comparable measurements. Nevertheless, the availability of lowcost digital multimeters and capacitance meters (sometimes combined in one instrument) does not negate the value of the basic bridge circuit.

The Wheatstone bridge

The Wheatstone bridge is the best known bridge circuit and the one most widely taught in electronics courses. It consists of two parallel resistance branches with each branch containing two series elements, typically resistors. A DC voltage source is connected across a diamond-shaped network to provide a source of current through the network. A *null detector*, originally a simple galvanometer, is connected between the parallel branches to detect a condition of balance.

The circuit shown in Fig. 1 was first invented by S.H. Christie in 1833. However, its value was not recognized until 1847 when Sir Charles Whetstone described how it could make accurate electrical measurements. As a result of his insights and demonstrations, the bridge came to be known as the Wheatstone bridge.

The Wheatstone bridge has probably been in use longer than any other electrical measuring instrument. Although still an accurate and reliable instrument, it is not as convenient to use as the new digital multimeters. Nevertheless, 0.1% accuracy is commonly obtainable with the Wheatstone bridge. This compares with the 3% to 5% error expected in resistance measurements made with analog ohmmeters. Accuracy on digital multimeters will depend on the meter's



FIG. 1—ORIGINAL WHEATSTONE bridge as demonstrated by Sir Charles Wheatstone in 1843.



FIG, 2—CONTEMPORARY VERSION of the Wheatstone bridge.

characteristics, such as DC voltage accuracy and resolution.

When making measurements with a Wheatstone bridge to determine the value of an unknown resistor (R_x) one of the remaining resistors is varied until the current through the null detector decreases to zero. The bridge is then in a balanced condition. This means that the voltage across resistor R3 is equal to the voltage drop across the unknown resistor R_x and the two divider resistors R1 and R2 are passing equal voltages. As a result, the following is true: (1) I1 × R1 = I_x × R_x

TABLE 1 RESPONSE OF 6-RANGE WHEATSTONE BRIDGE

Switch S1 Range	Bridge Range (Ohms)	Resistor R1 Value (Ohms)	R1/R2 Ratio (Nominal)	Bridge Null Sensitivity	
1	0-10	10	1/1000	3.0 %	
2	0-100	100	1/100	0.3 %	
3	0-1k	1k	1/10	0.03 %	
4	0-10k	10k	1/1	0.003 %	
5	0-100k	100k	10/1	0.03 %	
6	0-1MEG	1MEG	100/1	0.3 %	

At balance the voltage drops across R2 and R3 are equal so: (2) $I2 \times R2 = I3 \times R3$

Because no current flows through the galvanometer G when the bridge is balanced, it follows that:

(3) I2 = I1 and $I3 = I_X$

Substituting I2 for I1 and I3 for I_x in equation 1 yields the following equation:

(4) $I2 \times R1 = I3 \times R_X$

Dividing equation 2 by equation 4, yields.

(5) $R2/R1 = R3/R_{X}$

This can be rewritten as:

(6) $R2 \times R_x = R3 \times R1$

Equation 6 states the conditions of balance of a Wheatstone bridge and is useful for computing the value of an unknown resistor, once balance has been achieved.

Dividing both sides by R1 yields the most common Wheatstone bridge equation:

 $R_x = R3 \times R1/R2$

An important feature of the original Wheatstone bridge was its very high null sensitivity. Thus, if the bridge is energized



FIG. 3—WHEATSTONE BRIDGE with six DC measuring ranges.

from 10-volt DC, 5 volts is developed across all resistors at balance, and the galvanometer remains centered. A shift of only 0.1% will then give a 5 millivolt reading on a calibrated movingcoil analog meter. By installing a simple null-detecting DC amplifier, this circuit can have a *null sensitivity* factor (i.e., percentage out-of-balance detection value) of about 0.003%.

However, the major drawback of this 1847 bridge is that R3 must have a large resistive range if it is to balance all possible values of R_x . In 1848, the German-born engineer Wilhelm von Siemens, overcame this drawback by introducing the modifications, as shown in Fig.2

The bridge can be further modified as shown in Fig. 3 by giving R2 a fixed value and making R1 switch-selectable. This version is based on a 1970 version, high-accuracy laboratory measuring instrument.

The circuit in Fig. 3 can measure DC resistances from nearzero to 1 megohm in six switchselected decade ranges. Resistor R3 is a calibrated 10-kilohm variable potentiometer that controls the sensitivity of the balance-detecting centerzero meter. The value of RL limits bridge current to a few milliamperes. Table 1 lists the major weakness of this 1970 Wheatstone bridge-its null sensitivity (which is proportional to the R_x test voltage) degrades in proportion to the R1/R2 ratio's divergence from unity.

Consequently, the sensitivity is nominally 0.003% on the 10kilohm range where the R1/R2 ratio is 1/1, but it degrades to *Continued on page 145*



TELCO IN A BOX

Simulate a telephone line with the Telephone Company in a Box.

YOU JUST BOUGHT A BRAND NEW ANswering machine and you want to test it out. You set it up, connect it to the telephone line, and then just sit there looking at it. You want to see and hear it work, so you drive to a pay telephone and call your home number—nothing happens. Then you rush back home, decide to read the user's manual to find the problem, and then rush back to the telephone booth to test it again.

Now suppose that instead of an answering machine, you have some old telephones lying around in your house that don't work any more. To test those telephones completely, they must be connected to your telephone line and then someone must call you. Perhaps you have a modem or a fax machine that you would like to test but don't want to pay your local copy shop \$1 a page.

JAMES E. CICON

These hypothetical examples are some of reasons why you will want to build a Telephone Company in a Box, or TCB for short, the subject of this article. It will solve all of the problems posed in these scenarios, cheaply and quickly. All of the components needed to build this project are readily available. You will not need a PC board, nor will you be required to program a microprocessor or microcontroller.

The telephone company

To understand how the TCB works, it is helpful to know what happens on your telephone line when an outgoing call is made or an incoming call is received. When a telephone handset is on-hook, the telephone line voltage is about 50 volts DC. When a handset is offhook, the telephone line is loaded and the voltage drops to about 7 volts DC; this voltage is detected by the telephone company's central office equipment as an off-hook condition. The central office then provides a dial tone and the equipment waits for you to start dialing. When you dial a telephone number, the central-office equipment halts the dial tone, waits to receive a valid dialed number, and then makes a connection between your telephone and the number you dialed.

When the telephone company's central office equipment has a call for you, it rings your telephone by pulsing the 50-volt DC line voltage on and off at about 20 Hz for a short period of time, then it pauses, pulses again, pauses again, and so on. This produces the ring-ringring effect that informs you that there is an incoming call. When



FIG. 1—TCB SCHEMATIC. When telephones plugged into J1 and J2 are both off-hook, they transmit and receive their own audio.

you pick up the handset, the telephone line voltage drops back down to about 7-volts DC. The central office equipment detects the drop in line voltage, stops ringing the telephone, and connects your telephone to the calling party.

How does TCB work?

Two telephones can be plugged into the TCB, which will then simulate all telephoneline functions. When both telephones are on-hook, the TCB supplies 24-volt DC line voltage to both telephones. Although the telephone company supplies 50 volts, the 24 volts will work well because the resistors in series with the telephones have lower values than those used by the telephone company. Moreover, 24 volts DC is both safer to work with and easier to generate.

When one telephone handset is taken off-hook, its line voltage drops to 7 volts DC. The TCB senses this and, if the other telephone handset is onhook, it emits a ring signal. The ring signal is a 60-Hz sinewave at about 37-volts peak-to-peak. It is applied to the line for one second and halted for one second, repeatedly, until the other handset is picked up. Although the telephone company's ring signal is a 20-Hz squarewave at 50-volts peak-to-peak, the lower ringing voltage is used for two reasons. One is for safety, and the other is because it's easier to pick the ring voltage directly from the secondary of an AC transformer than to generate it with additional circuitry.

When both handsets are offhook, both lines are loaded to 10-volts DC. The TCB senses the condition, halts the ring signal, and connects both telephones together. Note that when the TCB is ringing the on-hook telephone, it is also sending the ring signal to the off-hook telephone. Depending on the design of the off-hook telephone, it might or might not ring. You might also hear the ringing from the speaker of the handset as a loud buzzing. This is not a problem—just a distraction. Telephones are designed to accept a ring signal even when they are off-hook.

Circuit description

Figure 1 is the schematic for the TCB circuit. When both handsets are on-hook, resistors R1 and R2 supply power to them, and the rest of the circuit can be ignored. When both telephones are off-hook, they transmit and receive their own audio. so all that the TCB does is supply power to them through R1 and R2. If switch S1 is closed, resistor R6 simulates a telephone being plugged into J2, permitting the testing of only one telephone. The rest of the circuit can be ignored in this example.

Now consider the situation where one handset is on-hook and one is off-hook. That causes

PARTS LIST

All resistors are 1/2-watt, 5%. R1, R2-560 ohms R10-3900 ohms H3-2200 ohms R4, R7-27,000 ohms R5-270,000 ohms R6-330 ohms R8-1800 ohms R9-220 ohms Capacitors C1-10 µF, 35 volts, electrolytic -22 µF, 35 volts, electrolytic C2-C3-470 µF, 35 volts, electrolytic Semiconductors D1-1N4001 diode Q1, Q2-2N3904 NPN transistor LED1-light-emitting diode, any color IC1-LM317 3-terminal adjustable positive voltage regulator SCR1-200-volt, 6-ampere silicon controlled rectifier (Radio Shack No. 276-1067) Other components F1-1/2-ampere fuse J1, J2-RJ-11 telephone jack (see text) PL1—AC plug and linecord RY1-SPST reed relay (12 VDC, 11 mA coil; 1-ampere, 125 VAC contacts) RY2-DPDT miniature relay (12 VDC, 43 mA coil; 1-ampere, 125 VAC contacts) S1, S2-SPST switch T1-120/25.2 VAC transformer, 450 mA Miscellaneous: perforated con-

Miscellaneous: perforated construction board, fuse holder, project case, wire, solder the on-hook line to go to 24 volts DC and the off-hook line to go to 7 volts DC. The coil of relay RY1 is connected across the two telephone lines, and the voltage difference between the two lines energizes it. When the contacts of RY1 are closed, C2 charges through R4; it takes about 1 second for C2 to charge to 12 volts DC. The 12-volts DC across C2 Resistor R8 supplies gate current to the SCR, which is organized so that C2 must charge to 12 volts DC before the SCR will turn on. When SCR1 turns on, the switch is essentially closed, and relay RY2 is energized. When the voltage across C2 drops to about 2.4 volts, Q1 and Q2 force SCR1 to turn off, thus turning the electronic switch



FIG. 2---THE FINISHED BOARD can be mounted in a metal case. Jacks J1 and J2, switches S1 and S2, and LED1 are mounted on the front panel.

causes a voltage-controlled switch consisting of R5, Q1, Q2, SCR1, and R8 to close, thus energizing RY2.

When RY2 is energized, RY1 is removed from the circuit and a 60-Hz, 37-volt peak-to-peak sinewave is placed on the telephone lines causing the telephones to ring. Because RY1 is removed from the circuit, capacitor C2 starts discharging through R7. It takes about 1 second for the capacitor to discharge to about 2.4 volts DC. That lower voltage level causes the voltage-controlled switch to disable RY2, removing the ring voltage from the telephone lines and putting relay RY1 back in the circuit. If one telephone is still off-hook and one is onhook, the cycle is repeated.

The voltage-controlled switch works as follows: Transistors Q1 and Q2 are connected as a Darlington pair, which couples the voltage across C2 to SCR1. off. Resistor R3 and LED1 provide a power-on indication for the user.

Construction

The circuit is so simple that a PC board is not necessary. Use point-to-point wiring. The power cord and AC plug PL1 were obtained from a 6 foot extension cord that usually sells for less than \$1. The female end was cut off, making it a linecord. Jacks J1 and J2 were two parts of a duplex telephone jack adapter that was cut apart. Snap off the male end and solder the wires directly to the exposed leads. However, any RJ-11 jacks will work well. Mount the finished board in a metal case. Any standard project case will do. Mount jacks J1 and J2, switches S1 and S2, and power indicator LED1 on the front panel of the case. Figure 2 shows the inside of the author's prototype unit.

How to use TCB

The TCB is easy to use. Plug a telephone into each of the two telephone jacks and turn the unit on. If you pick up the handset of one of the telephones, the other telephone will ring. When the handset of the other telephone is picked up the ringing stops, and you can talk normally between the telephones, as if a connection had been made by your local telephone company.

If you are testing an answering machine, plug it into one jack, and plug a standard telephone into the other jack. Pick up the handset on the standard telephone, and the answering machine will start ringing and should pick up the line. You will then hear the answering machine's outgoing message in the handset you are holding. The same procedure applies to the testing of modems, fax machines, or virtually any other telephone device whose operation you want to verify. Ω



An "all-channels" FM transmitter

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"All channels" FM transmitter

I've had several helpline requests for this one, as a paging override for a factory or plant cable system, or as a "please buy this house" realtor's message one that is received on an FM tuner set to *any* station. I can also see several emergency vehicle traffic-clearing applications for this as well.

I don't have a plug-and-go answer, but I do have a fairly simple paper design that *might* work, one that uses more of that Fourier series I have been looking at. Any decent FM receiver will exhibit the capture effect: When two competing signals are being received, and if one of them is even slightly larger, then only the stronger signal shows up as output audio, interference free. The capture effect happens because of the receiver's limiter circuitry. So, that's a good starting point. If your nearby signal is stronger than a distant one,



FIG. 1–A FAST RISE 100 KHZ SQUARE WAVE has harmonics well into the FM band. The harmonics match existing FM station channels.

Electronics Now, September 1995



FIG. 2 - ONE POSSIBLE SCHEME for an "all channels" FM generator.

then only the stronger nearby signal will be received.

I've written about stereo broadcasters a few times in past columns. The best starting point is the *Rohm* BA1404 stereo chip. The BA1404 makes a dandy low-cost stereo multiplexer, but is a terrible transmitter. Frequency drift and tuning accuracy slashes range or outright stops reception on digitally synthesized receivers, especially with car radios.

The one- or two-channel solution is to use the BA1404 to drive a crystal-stabilized FM transmitter. But linear frequency modulation of a crystal oscillator is more than a little bit tricky. A lab full of obscure test equipment, a lot of math, and years of experience are needed before you can even think about doing it.

I've explained how the *Pioneer* CD-FM-1 or the *Sony* XA7A auto CD adapters give you an off-the-shelf design solution. But the real problem here is not to transmit on a single channel, but to create *identical* signals on *all* the FM channels at once. Those FM channels run from 88.1 through 107.9 MHz, with an 0.2-MHz channel spacing.

Figures 1 and 2 show how the Fourier Series "comb transmitter" approach might handle this for us. A square wave has a fundamental frequency plus 1/3 the third harmonic, 1/5 of the fifth harmonic, and so on. So, start with a single-channel transmitter using one of those CD car adapters. Divide that down to precisely 100 kHz, making sure the final stage is an ultra-high-speed divider with a rise time better than two nanoseconds.

At this point, you have an FM square wave with lots of harmonics. For instance, there will still be two millivolts or so of the 881st harmonic at, of all places, 88.1 MHz. And, surprise, surprise, there is also an 883rd harmonic at 88.3 MHz, an 885th harmonic at 88.5 MHz, all the way up to the 1079th harmonic at, wonder of wonders, 107.9 MHz. Sure, the highend "stations" will be a tad weaker than the low-end ones, by about 20 percent or two decibels or, from the center of the band, plus or minus one percent, or one decibel either way. So, just add some gain and a bandpass filter to build up an all-channel FM broadcaster. Those one-dollar MAR chips from *Mini-Circuits Labs* should work just fine for predrivers in the circuit.

To be sure, you'd have problems getting this one to behave in the real world. A wide dynamic range would be needed, as well as thorough shielding. To prevent instabilities, you might want to mill each stage out of a block of solid aluminum. Or, at the very least, use stripline and double ground planes all the way around. There's also some very restrictive FCC regulations on this sort of thing. Transmitting unwanted messages to anyone, ever, is a definite no-no. But neither of these should be a problem on any private cable system. One solution to a "buy my house" application is to use several ultra-low-power curbside sites. Or perhaps go to some sort of a "lossy line" leaky cable that acts as a distributed antenna.

Filter fundamentals

An electronic filter can be defined as a frequency-selective network. A filter can emphasize wanted information while rejecting noise. *Low-pass*

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FIG. 3 - COMB FILTER BUILT UP by using switched capacitor techniques. Lowpass, bandpass, and highpass versions are also possible.

filters sustain the lower frequencies but obstruct higher ones. One example of a low-pass filter is the treble control on any hi-fi. Noise-reducing capacitors connected to ground in a power supply line are also low-pass filters, often called *bypasses*.

Bandpass filters select midrange frequencies while rejecting the higher and lower ones-the tuning dial on a radio, for instance. Audio equalizers are made from overlapping bandpass filter circuits, each of which can be adjusted independently.

Higb-pass filters block lower frequencies but emphasize the higher ones. The bass control on a hi-fi is one example of a high-pass filter. By adjusting its cutoff frequency, you can selectively de-emphasize any of the lower musical notes.

Actually, there is no such thing as an electronic high-pass filter. One of these would also be able to transmit microwaves, heat, light, and cosmic rays as well. The frequency response of any electronic circuit usually sets an upper limit, making all high-pass filters into bandpass ones.

Similarly, a true low-pass filter must respond down to DC, including any circuit bias or offsets. Audio lowpass filters add a large decoupling capacitor to block DC, and thus become a bandpass filter in the process. Other more specialized filter types include *band-stop* filters that reject a range of frequencies, *notch* filters that are just very narrow bandstop filters, and *delay networks* that adjust phasing at chosen frequencies.

There are several ways to build a filter. Classic filters use fixed inductors and capacitors. Transmission-line filters are just the UHF or microwave equivalent of classic L-C filters. The big difference is that the inductance and capacitance is distributed, rather than lumped. Active filters use combinations of resistors, capacitors, and op-amps instead. Advantages include low cost, high gain, and easy tuning.

Mechanical filters make use of the physical resonance of a crystal, ceramic resonator, or magnetostrictive rods. Surface-wave filters use constructive and destructive interference of sonic waves on a piezoelectric surface. A *switched-capacitor* filter is a hybrid analog and digital scheme-more on this shortly. Finally, *digital* filters will treat the filter problem as a group of numbers with digital signal processing, FFT transforms, or even wavelets.

The *order* of any filter determines how strong or how powerful it is. For instance, a *first order* low-pass filter should, at best, fall off at six decibels per octave, cutting its response in half as the frequency doubles.

A second-order low-pass filter can have a response that drops 12 decibels per octave. An eighth-order filter drops down at a rate of 48 decibels per octave. This gets real tricky on bandpass designs. For instance, a secondorder bandpass filter might drop off rather steeply just outside its passband, but its ultimate high- or lowfrequency falloff will be six dB per octave. The initial falloff rate is set by the Q of the filter which, for a second-order section, ends up as the inverse of the bandwidth.

The style of a filter is set by the underlying math used to determine its response. Filters using Bessel math provide the flattest time delay but a poor amplitude response. Butterworth filters provide the flattest passband amplitude. Chebycheff filters provide faster amplitude drop-offs, but at the price of lumpy passband ripple. Elliptic filters will add one or more notches just into the stopband, and give the steepest possible falloff, but these suffer from far poorer ultimate signal rejection.

Actually, these strange names are part of a *continuum* of possible filter responses. In an active filter, all you have is a stack of cascaded first- and second-order sections. Each one has a *cutoff frequency* and a *damping* factor. By suitably changing these values, you can create any filter from Bessel on up through elliptic.

In general, you can't do nearly as much with a filter as you'd like to. Increasing the order adds cost, noise, linearity, dynamic range, shielding, losses, tuning, stability, and response problems, besides totally trashing the transients and time delays.

One nasty filter response is called *group* delay. For instance, if a modem tone frequency for a one has a different group delay than that for the zero, then at times you'll get a one, a zero, neither, or both.

In general, digital filters let you do much more, much better than analog filters. For instance, a filter with a "brick wall" or "reject everything" response is impossible with analog designs, but can be created easily with digital techniques. Group delay is also more easily dealt with, and so are the elegant new *constant phase* designs.

Sadly, digital filters are currently restricted to operating on big amplitudes and low frequencies, or to non real-time uses. And they have their own set of problems, not the least of which is *aliasing*, or generating nasty in-band artifacts from out-of-band signals.

By using wavelets or fast Fourier transforms, you can find the full spectrum of a waveform rapidly. Analog filters must instead be swept slowly. You can soon expect to see a revolution in radio receivers. Start at the front end, downconvert, and then digitize. From there, do everything digitally. Several technical notes on this are offered by Harris Semiconductor and Analog Devices. You'll find much more on filtering in my Active Filter Cookbook, now in its brand new second edition and seventeenth printing. See my nearby Synergetics ad for information on its availability.

Switched-capacitor filters

There's an interesting hybrid filter scheme based on both analog and digital technology. This is known as a switched-capacitor filter. The basic idea is shown in Figs. 3 and 4. The switch goes 'round and 'round at a clock rate that is a multiple of the desired center frequency. Now, for a given switch position, there's only one capacitor in the circuit. This capacitor should either increase or decrease its charge according to the difference between itself and the input signal. Each capacitor thus retains a sample of the recent previous input history.

Now for the neat part: At DC or

very low frequencies, the capacitors will pretty much track the input. As the frequency increases, the capacitors may charge or discharge. At a frequency of exactly *one-balf* the onetrip-around rate, any capacitor will charge on one pass and discharge on the next, giving a zero output. At a frequency of precisely the one-triparound rate, any capacitor charges identically on each pass and gives you a *maximum* output.

This simple example forms a *comb* filter. The first null frequency will be

determined by the switching rate and the number of capacitors. In this example, the first null frequency is at 1/16th the data rate. Comb filters can reject a fundamental and *all* of its harmonics. One major application for comb filters is separating NTSC color and brightness signals. Different combinations of resistors, switches, and capacitors could build high-quality low-pass, high-pass, band-pass, or band-stop responses. Switching is normally handled by CMOS analog switches.



Advantages of switched capacitor filters include precise tuning over a wide frequency range, plus lower cost and small size. And because everything is done on the silicon chip, there is no need for any external capacitors.

Dozens of switched capacitor filter chips are now available from several sources, some approaching a dollar each. Our resource sidebar for this month shows you several of the main suppliers involved. Of these, check into *Maxim* and *Linear Technology* first. Free samples of these filter chips are available from Maxim.

Sadly, there have been quite a few false starts involving these switched capacitor filters. Early versions had excessive noise floors, caused by the transient feedthrough of the clock. That happens because of the gate-tosource capacitance in any CMOS device. The dynamic range is set at the low end by the noise floor, and at the high end by clipping, distortion, or other nonlinearities. Today, these glitch and noise-floor problems have been minimized. Still, though, switched capacitor filters are strictly



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limited to larger signals. They are not at all suited for microvolt-size inputs. Typical devices are limited to ultrasonic or lower frequencies. But they sure are a useful tool.

A narrow bandpass filter

There is this interesting new LTC 1164-8 chip from Linear Technology. This device creates an easily tunable and ultra-narrow eighth-order elliptic bandpass filter. The bandwidth is one percent of the center frequency. Cost is around \$24 in single quantities. Figure 4 shows the details. The chip goes ahead of an op-amp in such a way that the generated noise floor is independent of the gain. With a single 5-volt supply, the current draw is around two milliamperes and the maximum center frequency is 4 kHz. The input clocking frequency is 100 times the center frequency. Output noise is half a millivolt. The stopband attenuation is fifty decibels or so. Careful ground-plane shielding is required. Be certain to carefully read the LTC1164 data sheet.

Gain is adjusted by changing the input resistor. Note that the *smaller* the resistor, the *higher* the gain. A gain of 1000 is possible, but you will get the best dynamics at unity gain. To use the circuit, set the gain by picking the input resistor, apply an input signal you want filtered, and a 100° square-wave clock derived from any old TTL or CMOS source.

For quick and dirty experiments, simply use one sixth of a 74HC14, a variable resistor, and a capacitor to make a square-wave clock source. One obvious use for the chip is for checking real-world harmonic performance of the magic sinewaves I looked at a few columns ago. But one major warning over any narrow band filter: If your signal is changing rapidly, most of it will be missed! That happens because the high Q of the filter takes a long time to build up a useful result. It's sort of the same as whapping a large pendulum with a hammer. Nothing much happens with one ping. Its only when hundreds of well timed pings occur that oscillation starts. Thus, narrow-band filters must be very accurate and tuned very slowly. And they will work only with very

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SWITCHED CAPACITOR FILTER RESOURCES

Analog Devices PO Box 9106 Norwood MA 02062 (617) 329-4700

Burr-Brown 6730 S Tucson Blvd Tucson AZ 85706 (602) 746-1111

Cermetek 1308 Borregas Ave Sunnyvale CA 94088 (408) 752-5000

Crystal Semiconductor PO Box 17847 Austin TX 78744 (800) 888-5016

Datel 11 Cabot Blvd Mansfield MA 02048 (508) 339-3000

EGG/Reticon 345 Potero Ave Sunnyvale CA 94086 (408) 738-4266

slowly changing signals.

The data sheet doesn't mention it, but I suspect that the LTC1164 also responds to the second and higher harmonics of your center frequency as well. This is called *aliasing*. Therefore some external prefiltering appears to be a good idea. This device seems ideal for a lot of telecommunication and midrange audio applications, especially for vibration studies and



FIG. 4–AN ULTRA SELECTIVE switched capacitor bandpass filter. The center frequency can be tuned by adjusting the digital $100 \times \text{clock}$.

Exar Corp PO Box 49007

PO Box 49007 San Jose CA 95161 (408) 732-7970

Linear Technology 1630 McCarthy Blvd Milpitas CA 95035 (408) 432-1900

Maxim 120 San Gabriel Dr Sunnyvale CA 94086 (800) 998-8800

National Semiconductor 2900 Semiconductor Rd Santa Clara CA 95052 (800) 272-9959

SGS-Thomson

1000 E Bell Rd Phoenix AZ 85022 (602) 867-6259

Texas Instruments PO Box 809066 Dallas TX 75380 (800) 336-5236

harmonic analysis. But, for ultra-low frequencies (such as seismography or brainwave research), digital FFT or wavelet techniques are the only way to go these days.

Two contests

Let's have a pair of contests this month. Either (A) Add to our dialog

TV/VCR TECHNICIANS



G Chassis VCR Repair

This video will take the fear out of servicing the G chassis, which is found in nost VCRs made from '92 to today. Displays how to easily align the mechanism, and also repair of many common G failures. 30 minutes, VHS. Order# VID17. \$39.95

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Other available video titles:

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- > Layman's guide to VCR Cleaning
- > Minor VHS Camcorder Repair
- > Nintendo Repair & Maintenance
- > Microwave Oven Repair
- > PC Computer Repair
- > PC Computer Assembly & Configuration



VCR/TV Repair TECHDISK

This software database contains over 4800 VCR & 2100 TV problems, along with practical repair suggestions for each, compiled from case histories. User friendly and menu driven. Add/delete your own problems too. Annual updates available for a nominal charge. (IBM) Order# 56-130. \$69.95 2 disk set. Also available separately for \$39.95 ea.

VCR Model SuperCross

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35 Windsor Drive Amherst NH 03031 (603) 429-0948

GEnie

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Harris Semiconductor Box 883 Melbourne FL 32902

(407) 724-7000

IC Card Systems & Design

6151 Powers Ferry Road NW Atlanta GA 30339 (404) 955-2500

Mini-Circuits PO Box 350166 Brooklyn NY 11235 (718) 934-4500

Nomadness Report PO Box 2185 El Segundo CA 90245 (310) 322-1655

Pioneer Box 1720 Long Beach CA 90801 (213) 835-6177

on "all channels" FM schemes, or (B) Show me an unusual new use for an ultra narrow, switched capacitor bandpass filter.

There'll be a dozen or so of my Incredible Secret Money Machine II books going to the better entries. Plus an all-expense-paid (FOB Thatcher, AZ) tinaja quest for two going to the very best of all. Please send your written entries to me here

at Synergetics, rather than to *Electronics Now* editorial.

New tech lit

An update on those active noisecancellation topics we looked into several columns back: I've uploaded a tutorial and resource directory as NOISCNCL.TXT. There's also a new Active Sound and Vibration Control News with free sample copies available.

More on FM radio RBDS services: There is now an Internet forum up at 56 http://pcbf1131e.util.ch/~uer/rdsh000

Quatech

662 Wolf Ledges Parkway Akron OH 44311 (800) 553-1170

Radio World

5827 Columbia Pky #310 Falls Church VA 22041 (703) 998-7600

Roberts Hydraulics

311 North Morgan Street Chicago IL 60607 (312) 829-1365

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Sony

9 West 57th St 43rd Fl New York NY 10019 (212) 371-5800

Sycard Technology 1180-F Miraloma Way Sunnyvale CA 94086 (408) 749-0130

.htm. As always, *Radio World* is the best place to get RBDS information. Hundreds of stations now provide RBDS services, ranging from song title and artist, traffic and weather, coupon promotions, on up to GPS navigation corrections.

Those new plug-in PCMCIA cards for laptop computers are also known as PC cards. One trade journal that zeros in on these is *IC Card Systems & Design*. A pricey source for

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US Internet email access link: SYNERGETICS@GENIE.GEIS.COM. books on the PCMCIA standard is *Sycard*. One source of cards is *Quatech*.

Bargain hydraulics are offered by *Roberts*. Unusual modelmaking parts are available at *Special Shapes*.

The magic sinewave stuff that I looked at a few columns or ago has now been improved and upgraded as MAGICSIN.PDF. I've got actual chips and working hardware. Codeveloper packages are newly available for this emerging opportunity.

For most small-scale startups most of the time, patents are certain to end up as a net loss of time, energy, money, and sanity. Find out why this is true in my *Case Against Patents* resource package. An autographed copy of my *Incredible Secret Money Machine II* is included in the patents resource pack.

My usual reminder that most of the resources mentioned appear in the Names & Numbers or that Switched Capacitor Filter Resources sidebar. I've managed to wrangle a special ten free hour GEnie PSRT trial per the Need Help? box.



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Conducto-Lube

From the report:

...compare the conducting properties of Conducto-Lube lubricant with factory lubrication and non-lubricated connections on an air switch. Identical test setups and procedures were used for each test so that comparative data could be collected and the relative performance of each type of connection could be quantified.

"The test results indicated that the contact resistance of the switch employing Conducto-Lube was generally lower than that of the factory lubricated switch and the switch that used no lubricant."

Cool-Amp How it works:

 Applies on the iob. Application is simple. Yet Cool-Amp adheres permanently. As tests show, it is better than electroplating. Minimizes overheating and power loss by silver plating high amperage connections. · Saves time, reduces maintenance. Cool-Amp is so simple to apply on the job. It assures maximum conductivity for copper, brass, or bronze contacts and prevents losses due to oxidation.



A one-pound bottle of Cool-Amp silver plating powder plates approximately 6,000 square inches.

Conductivity is demonstrated

by inserting test prongs into a container of Conducto-Lube and establishing a circuit. Photo shows low voltage (115 VAC) continuity through container.

Conducto-Lube How it works:

. This is the conductive lubricant; highly conductive because it contains pure silver.

· Originally developed to lubricate switches, to the point tension can be adjusted to factory specs allowing full rated capacity of the switch to be maintained at all times.

· Uses have continued to expland-from switches and breakers-to any application where a conductive lubricant is needed.

¹Various tests were performed on both products in the Electro-Test, inc. tacilities in Portland, Dregon during January-March, 1994. Levaluation of plating thickness of Cool-Arng was performed by Surface Science Laboratori of Mountain View, California.

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a	Kit: \$ 165.00 Asmb.\$ 195.00	ohms. Power Requirement: ±55 to ±65VDC 8A.Suggested Mark V model 009 Transformer. Capacitor 10,000uf 80V model 016. Recommended Metal Cabinet LG-1925.	**008 28/30V x 2 6A 40.00 **009 48/53V x 2 8A 66.00 **012 33/40/42V x 2 6A 48.00
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t	Kit:8 54 00 Asmb 8 74 00	V model 001 or 008 transformer. Capacitor 10,000uf 80V model 016. Recommended Metal Cabinet LG-1924.	50V. Current limit trip is adjustable up to max of t 3A. Suggested Mark V 002 transformer. (1 lb.) o
t		S A FET DYNAMIC BUFFER STEREO PRE-AMP. (1 Ib.)	SM-302 AA 60+60W STEREO POWER AMP. *
a I	SN1/1-2	Frequency Response (at rated output): Overall 10HZ-	It provides 3 input jack
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\$	a finite of	better than 90dB. Input Sensitivity (1KHZ for rated output): 300-800mV. Maximum Output Level: Pre-Amp	low level input sources. Power Output: 60W \$
9 9	Kit:#64.00 Asmb.# 80.00	output 1.8V (0.1% THD). Power Requirement: 30V X 2 AC 500mA.	per channel into 4 ohms RMS. 20HZ-20KHZ. 9 THD:<0.1%. input Sensitivity: Mic/Guitar 9
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template, paint mask, and detailed installation instructions. **Specifications:** 6-1/2" polypropylene cone woofer with poly foam sur-round. 41" textile dome tweeter/midrange. 48 ohm impedance. 43 com-ponent L/C crossover network. Frequency response: 50-20,000 Hz. •Power handling capability: 60 watts RMS/100 watts max. •Sensitivity: 89 dB 1W/1M. •Overall dimensions: 8-1/2" W x 12" L x 3-1/2" D. •Hole size: 7-1/4" x 10-3/4". •Fits into standard 2" x 4" wall. •Net weight: 12 be per pair. lbs, per pair.

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RATES

Electronics Now, September 1995

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If you use a Box number you must include your permanent address and phone number for our files. ADS SUBMITTED WITHOUT THIS INFORMATION WILL NOT BE ACCEPTED.

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Ads not received by our closing date will run in the next issue. For example, ads received by November 13 will appear in the March issue that is on sale January 17. ELECTRONICS NOW is published monthly. No cancellations permitted after the closing date. No copy changes can be made after we have typeset your ad. NO REFUNDS, advertising credit only. No phone orders.

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9 - \$37.50 10 - \$37.50	11 - \$37.50	12 - \$37.50	37 - \$92.50	38 - \$95.00 39	- \$97.50 40	\$100.00	
13 \$37.50 14 - \$37.50	15 - \$37,50	16 - \$40.00	Total words \$		\$2.50 per word = \$		
17 - \$42.50 18 - \$45.00	19 - \$ 47.50	20 - \$50.00	Bold Face \$0		\$0.40 per word	= \$	
21 - \$52.50 22 - \$55.00	23 - \$57,50	24 - \$60.00	Special Heading \$		\$30.00 =	\$	
25 - \$ 62.50 26 - \$65.00	27 - \$67.50	28 - \$70.00	C	Other	=	\$	
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SE /200

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Tektronix 496P Programmable Spectrum Analyzer

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BY JEFF HOLTZMAN

How to solve all your PC problems

Go cold turkey: sell your PC.

ONUNDRUM: HOW CAN THE FOL-

LOWING BE TRUE: PC = PC? IN

HONOR OF THE LAST MONTH OF SUMMER VACATION, I'D LIKE TO DEDICATE THIS MONTH'S COLUMN TO A VERY

serious subject, albeit one with a lighthearted side. The subject concerns the very difficult problem of personal-computer ease of use. In spite of all our best efforts during the past few decades, things are not getting better, but worse. Following are several ideas for dealing with the problem in an efficient manner. These ideas are descriptions of several magazine articles I don't have time to write, but would love to read. (Note placement of tongue firmly in cheek.) The articles are:

•Why the computer will never get easier

•WYSIWYG is really WYGIN-WYS

•Real men prefer the command line interface

•Sell your copy of Windows, put your kids through college

- Networks are for nitwits
- Do what Nancy does

•Renounce graphics, sound, and video

•Forswear Microsoft products and survive

•Go cold turkey: sell your PC

Why the computer will never get easier

The computer industry has a vested interest in difficulty. If it was easy, anybody could do it, and if anybody could do it, lots of people would lose their jobs. The subterfuge then is to proclaim a goal that is the exact opposite of the real goal, and to keep shifting the focus to various pieces of the puzzle. Doing so will ensure that no one, neither customers nor competitors, will ever be able to figure out what the goal is, or even if there is one.

Approximately every five years or so, propose a new technology as



FIG. 1-RESOLVE ALL YOUR PC PROBLEMS in one fell swoop. Just say no.

the universal solution to the problems facing computers. Examples include Structured Programming, Networks, Artificial Intelligence, and Object-Oriented Programming. For the past few years, the GUI has been touted as the solution to difficulty. Yet has no one noticed that the complexity of system configuration has increased geometrically? Do you know what every line in your CONFIG.SYS about does? How your SYSTEM.INI? Do you think that is an accident?

WYSIWYG is really WYGINWYS

Pronounce that as "Whizzy Wig is really Wiggin Whiz." The theme of this article is: "What You See Is What You Get" is a lie; the truth is, "What You Get Is Never What You See."

This article will not bore us with technical descriptions of device-independent programming models, resolution differences between printers and monitors, page-description languages, markup languages, or anything of the sort. The real purpose is to expose the collusion among major manufacturers and the government. In the name of full employment, the government has worked behind the scenes with industry leaders to ensure that products from different manufacturers are fundamentally incompatible with one another.

One of the primary beneficiaries of this economic system is the government itself, which has to set upsecretly, of course-incompatibility certification labs (ICLs). The purpose of these labs is to certify products as incompatible at a fundamental level. Manufacturers are forbidden from releasing commercial products that show any significant level of compatibility with similar products from other manufacturers. Running these labs is actually the purpose of the super-secret National Security Agency (NSA), which contrary to popular opinion measures security in terms of economic stability, not military might, stealth, or sophistication.

The software industry is the major beneficiary of the activities of the ICLs. Devising, developing, and supporting complex models that abstract common features helps ensure employment for a wide range of professionals: modelers, programmers, testers, technicians, management, third-party support organizations, and more. The bugs and inconsistencies thereby created help feed a press hungry to print bad news. It also feeds consumers' increasingly cynical views of both government and business. It also contributes to an everexpanding spiral of bug fixes and upgrades. Full employment for everyone! A chicken in every pot! A bug in every product!

Real men prefer the command line interface

(Women aren't really interested in technology, so don't worry about the implicit sexism in the title of this article.)

Whereas early articles showed that WYSIWYG is not all it's cracked up to be, this article begins laying the groundwork for an anti-WYSIWYG backlash. The upshot is this: Anything GU(I) can do, CLI can do better; CLI can do anything better than GU(I).

Using examples such as vi, TROFF and T_EX , this article will prove that CL-based edit/format duos can do anything that GUI-based word processors and desktop publishers can do, but at significantly less cost, with greater reliability, and with greater understanding on the part of the user.

A related topic is: "Who needs spreadsheets when you've got COBOL and FORTRAN?"

Sell your copy of Windows, put your kids through college

After proving the higher innate efficien~y and accuracy of the CLI, the next step is to divest yourself of your GUI, most likely Windows. Right away you'll recover a good 40 MB of disk space. You'll also start sleeping better because you'll be able to start focusing on content once again, leaving appearance to someone else. You'll thrill to monospaced printouts in Courier. You'll revel in tools simple enough to operate without thought—like riding a bike. Productivity will zoom at least 400% within days! Of course, without Windows, you'll have no need for Windows applications. Depending on the degree to which you had been led down the path, you should be able to recoup another 100 MB of disk space, or even more. In the return to CLI computing, this essentially means that you'll never have to buy a computer again. So instead of perpetually buying products with 4 to 5 year obsolescence cycles, you can buy a computer once and bank the rest. Potential lifetime savings could easily amount to \$100,000 or more.

Networks are for nitwits

People who really want to share files trade floppies, temporarily connect using LapLink, or use an on-line service. Corporate networks are simply a way to usurp individual power. The primary beneficiaries are the network dweebs, not the users, and certainly not the corporation.

Even within the computer industry, network dweebs occupy an exalted position. They have their own jargon, incomprehensible even to dweebs in related specialties. They're responsible to no one, come and go as they please, and use any flimsy excuse to bring the network down at the most inopportune moment for users. Within a PC, nothing—not even multimedia—causes more trouble than networking hardware and software.

So to regain control over your life and simplify your existence, forego networking.

Do what Nancy does

Is it an accident that every 18 months or so a whole new generation of computer hardware is released that makes everyone want to dispose of their current inventory and upgrade? No, it is no accident. But that doesn't mean we have to respond like trained rats to some Skinnerian buzzer.

When it comes to hardware (and software) upgrades, take a tip from Nancy Reagan: Just say no.

Renounce graphics, sound, and video

Multimedia has already shown itself to be more concerned with sur- 125

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A public service of this publication and the Consumer Information Center of the U, S. General Services Administration face presentation than deep content. So giving up the CD-ROM and sound card should not be difficult. Doing so will result in further gains in hard-disk space. A side effect will be a two-order-of-magnitude reduction in system complexity, and a corresponding increase in robustness and reliability. No more unexplained crashes! No more frozen mouse pointers just before your file is actually saved.

Forswear Microsoft products and survive

Yes, this is a capitalist society, but enough is enough. The next step on the way back to a simpler, more selfdetermined way of life is to forswear use of all products made by Microsoft. We started by giving up Windows; now let's throw in DOS, networking, applications, programming tools, games, utilities, and anything else that may be lurking around. (We're considering offering a one-time-per-user exception for one copy of a crucial Microsoft program, such as QBasic or Flight Simulator.) Conservative estimates are that system stability will increase at least as much as by giving up multimedia. We also expect significant disk and memory savings as well.

Go cold turkey: sell your PC

I think my high-school class was among the last to learn how to use a slide rule. You probably don't even know what one is. (If not: A slide rule is a self-powered low-maintenance analog device for multiplying and dividing by adding and subtracting logarithms.) It was easy to get three places of decimal accuracy from a slide rule; whereas today's CPUs (e.g., Pentium) and applications (e.g., Windows calculators) can't even guarantee that. So you don't need a PC to do accurate number crunching.

Nor do you need one to do word processing. Used equipment stores are filled with typewriters rendered obsolete by the computer revolution; they can be had for a song. However, a typewriter can be a high-maintenance item just like a PC; you may want to simplify even further to what Byte magazine long ago referred to as the "Generic Word Processor:"the standard wood-barreled #2 pencil. An electric pencil sharpener is a politically correct peripheral, although favor will shine on those who opt for manual devices.

Ask yourself: Did Newton need a PC to invent calculus? Did Einstein need one to invent the theory of relativity?

Conundrum resolved

Politically correct equals no personal computer.

E-mail

Someone sent me a huge uuencoded file via internet mail. The file contained a demo of something the sender thought I (and you) might be interested in. I'll never know because I aborted the download as soon as I realized what was going on. Sending large unsolicited files via email is very bad etiquette. Feel free to send a short, concise summary of your product—but not the product itself—via email. If I'm interested, I'll get back to you.

Someone else suggested doing a column on Linux, a UNIX clone for Intel-based CPUs. In essence, Linux is a shareware product developed and supported by programmers all over the world. I have heard and read good things about Linux, but to date have no hands-on experience with it. Frankly, the thought of installing it and learning enough about it to say something intelligent is not terribly appealing, based primarily on years of difficult experiences installing and testing commercial operating systems, such as DOS, Windows, OS/2, and other Intel UNIX clones. On the other hand, I'm not totally against it, either. If there's a lot of interest, I'll give it a second thought. Let me know.

Also, I enjoyed hearing from 14year old Billy K., from Des Moines. Billy is interested in everything to do with computers and electronics, and I bet someday he'll turn out to be a crack engineer.

You can reach me on CompuServe as 72170,2226, or via the Internet at 72170.2226@CompuServe.Com.



Your HF Digital Companion:

Explore HF Digital Communication with Your Multimode Controller by Steve Ford, WB81MY. The American Radio Relay League 225 Main Street Newington, CT 06111 Phone: 203-666-1541 Fax: 203-665-7531 **\$10.00**



This book is an introduction to highfrequency (HF) communication that provides the practical information needed for the amateur radio o perator interested in

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communicating in that region of the spectrum. It is written simple English without being filled with confusing jargon.

Steve Ford explains the differences between HF digital communication and other operating modes and its advantages. He explains how to set up a HF digital station that permits the use the digital modes to talk to radio amateurs anywhere in the world.

The book explains such subjects as hunting for digital DX, work mailbox and bulletin-board systems, and how to compete in contests that test operating skills. Other topics discussed are RTTY and the AMTOR, PacTOR, CLOVER, and G-TOR digital modes. The book contains lists of helpful books and newsletters as well as sources for equipment and software. An extensive glossary defines the terms unique to communication in the digital modes.

E-Mail Security:

How to Keep Your Electronic Messages Private by Bruce Schneier: John Wiley & Sons, Inc. 605 Third Avenue New York, NY 10158-0012 Phone: 1-800-CALL-WILEY **\$24.95**

Today, mes-

sages can be

sent by com-

puter modem

a lot faster

than they can

by sent by the

traditional

services from

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system to Federal Express or UPS. However, in transmitting over open telephone lines, your message is far more vulnerable to interception by unwanted persons than ever before. So you have a tradeoff between speed and security.

This book is all about how to regain some or all of the privacy lost to E-mail. The average electronic mail message passes through a half dozen intermediate stages between its source and its destination, and there are no laws to prevent prying eyes from reading that message.

You probably don't want coworkers, hackers, or even your supervisor to read you messages. Worse yet they might be intercepted by business competitors or others intent on prying into your personal affairs. In the absence of laws governing the interception of E-mail (which perhaps could not really be enforced anyway), you are responsible for protecting your own computer communications.

Bruce Schneier tells you how to do so. He has divided his book into two sections. The first section is devoted to the issues of privacy, electronic mail, encryption, authentication, and keeping your private "key" private. The roles of patents, government, and export laws are examined.

The second section focuses on achieving E- mail privacy. It warns users about how messages are intercepted, read, copied, and even (in some cases) altered for malicious purposes. The author explains how you can avoid those breaches of privacy.

Mobile Communication Satellites:

Theory and Applications by Tom Logsdon. McGraw-Hill Book Company 11 West 19th Street New York, NY 10011 Phone: 1-800-2-MCGRAW \$55.00



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the constellations of communica-

tion satellites that are now being

This book is all about the latest advances in global wireless communication. Tom Logsdon explains the present status and the future prospects for

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deployed in geosynchronous, low-altitude, and medium-altitude orbits. He then describes how those satellites are revolutionizing communications and navigation for ships, aircraft, commercial and emergency vehicles, and even private cars.

The book includes explanations of such important communications concepts as constellation selection, spread-spectrum, digital implementation, and data-compression. You'll find out how analytical tools select the most efficient constellation configurations and how mobile communication satellites are designed. You'll also learn how and why signal formats and frequencies are selected.

Getting Started with Lotus 1-2-3 for Windows

by Barbara R. Farrell. John Wiley & Sons, Inc. 605 Third Avenue New York, NY 10158-0012 Phone: 1-800-CALL-WILEY **\$13.95**



Barbara Farrell has written this primer on Lotus 1-2-3 for Windows for use in introductory personal computing courses. The book is intended for those persons with a working

knowledge of personal computers and Windows but no previous experience with the Lotus 1-2-3 spreadsheet program.

Basic skills are taught in short segments of focused activities based on realistic applications. Each lesson has eight key elements to help the reader master important concepts and skills. The emphasis is placed on the reader's ability to apply those skills in daily work.

Learning objectives are given at the beginning of each lesson, and the book is organized so that the reader can meet those objectives while actu-

ally preparing spreadsheets. This 128 approach encourages learning by instructing the student in how to accomplish a practical objective and reinforcing that learning.

Lists of procedures provide ready references to the steps required to complete each lesson, and step-bystep instructions teach the reader how to apply those lessons to solve specific spreadsheet objectives.

The reader is encouraged to practice newly acquired skills by doing independent projects with decreasing levels of text support. A reference section at the end of the book lists all of the commands needed to complete the projects presented in the text.

Living with the Chip:

How the Chip Affects Your Business, Your Family, Your Home, Your Job, and Your Future by David Manners and Tsugio Makimoto Chapman and Hall One Penn Plaza New York, NY 10119 Phone: 212-564-1060 **\$16.95**



This book presents an essentially non-technical overview of the integrated circuit and the industry it spaw-ned. It describes how the devices work, reviews the history of their development, discusses their present stadictions about the

tus, and makes predictions about the future directions and trends for these semiconductor "chips."

The authors start off by discussing how microelectronic device advances have led to the development of ever more sophisticated, reliable and lower cost products such as radios, calculators, personal computers, fax machines, telephones, VCRs and camcorders.

The book than goes on to review the history of the IC, naming key contributors to its development and identifying and describing the roles of the many multinational manufacturers who have had or still hold key positions in the technology.

The concept of the "learning curve" is described, relating it to the normal decline in prices as production ramps up. You will also learn about the necessity for having second or alternative sources for popular devices.

The reader will find useful information here about the specialties of the various manufacturers that should be useful in tracking down sources for specific kinds of chips. The book even ventures into a discussion of corporate reputations, listing those manufacturers who have remained innovative and dynamic (making them attractive employers) and those which have fallen behind the flock.

"The Audio Source" Summer Sale Catalog

Sescom, Inc. 2100 Ward Drive, Henderson, NV 89015-4249 Phone: 800-634-3457 (orders only) or 702-565-3400 Fax: 800-551-2749 (orders only) or 702-565-4828





This Sescom catalog offers a range of audio products and accessories that are available at discount prices. The products include those designed to "plug-n-play" in existing systems. The recently

designed, easy-to-connect, easy-touse accessories can be installed in rack systems or placed in appropriate positions in the listening room.

The products include a four-channel audio distribution amplifier, a stereo-buffer amplifier, and a stereo combiner. The catalog describes a selection in-line devices that eliminate AC hum. These include XLR connector adapters, audio SIPs, portable microphone line drivers, a dual interface box, and a stereo line driver.

Digital Power Supplies Brochure

Kepco, Inc.

131-38 Sanford Avenue Flushing, NY 11352 Phone: 718-461-7000 Fax: 718-767-1102 Free



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This Kepco brochure describes the company's line of digitally programmable power supplies for integration into host test systems. The supplies make use of either the SCPI or CIIL control

languages and communicate with either the GPIB (IEEE 488.2) or a long-range serial bus (IEEE 1118); VXI is also supported.

Kepco has introduced its line of MST high-density, 200-watt programmable power supplies. The latest design concepts permit the package size to be reduced so that nine modu-

lar supplies will fit in a standard 19inch rack. Text and illustrations describe how to installing multiple power supply modules, each rated for up to 1000 watts.

Test Instruments Catalog

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in the selection of the right instrument for your requirements are included in this catalog.

Industrial Components Catalog

Mouser Electronics 958 N. Main St. Mansfield, TX 76063 (817) 483-4422 (800) 992-9943 Free



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tors, wire and cable, and connectors. The 316-page catalog contains specifications, drawings, and guaranteed prices for over 53,000 products.



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BY FRANKLIN J. MILLER

AUDIO UPDATE

AUDIO CONNECTOR Overview

XLR, Unbalanced, and DIN Connectors

HREE DIFFERENT CONNECTORS ARE IN WIDE USE TODAY IN CONSUMER AUDIO EQUIPMENT. THEY ARE THE SO-CALLED RCA PLUG AND JACK, THE ¹/4-INCH MONO AND STEREO PLUG AND

jack and the 3.5-millimeter mono and stereo plugs and jacks. DIN and XLR connectors are more likely to be found on top-of-theline consumer audio products and professional audio equipment for recording studios and radio and TV stations.

The connector specified for audio equipment depends on the price and aqpplication of the audio equipment.

Regardless of where the equipment is made, the manufacturer must choose standard connectors that are compatible with and appropriate for the price and application of the equipment in the country where he intends to sell it.

The mini plug accepts wire with shield braid. The tip of the plug contacts the center conductor of the wire, while the cylindrical metal extension of barrel of the plug contacts the shield braid. The wire has some of the properties of unbalanced coaxial cable, but the shield braid is not coaxial with the central lead. If the barrel has internal threads, it is identified by its diameter and the number of threads per inch. A $\frac{5}{8}$ -inch diameter barrel with 27 threads to the inch would be specified as ${}^{5}_{8} \times 27$ TPI.

The phone plug is a larger version of the mini plug and is used to terminate cables joining audio components. The phone plug looks like the mini plug but it is called a phone plug because it was originally used as a connector for manually operated telephone switchboards. Phone plugs can have two or three elements, and they have barrels that are $1\frac{1}{4}$ -inch (31.8 mm) long.

Phone plugs are used primarily on headphones and microphone



FIG. 1—PHONE PLUGS: The mono, two-conductor version (a) and the stereo, three-conductor version (b).

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FIG. 2-PHONE JACK: 1/4-inch mono jack (a) and its schematic symbol (b).

cords. They are available in $\frac{1}{4}$ -inch (6.35 mm), $\frac{1}{8}$ -inch (3.5 mm) and $\frac{3}{32}$ inch (2.5 mm) plug diameters. In those parts of the world where metric measurements are standard, the $\frac{1}{8}$ inch (0.141-inch diameter) plug is called the 3.5 mm plug. Both twoand three-conductor styles with matching sockets called jacks are readily available.

Figure 1-a shows a monophonic phone plug. An insulated ring isolates the *tip* from the *sleeve* to provide two insulated contact surfaces. The dotted lines represent the internal wiring. Figure 1-b shows a stereophonic plug that also has an insulated tip and sleeve, but a third insulated element called the *ring* is inserted between the tip and sleeve to provide a third insulated contact surface. Both mini and phone plugs can be either monophonic or stereophonic.

In both the 1/4-inch and 3.5 mm mono and stereo plugs, the sleeve is the ground connection and the tip is the "hot" connection. In the 1/4-inch and 3.5-mm stereo plugs, the ring is the "cold" connection for left or outof-phase signals and the tip is for right or in-phase signals. Phone plugs are available in right-angle form.

Figure 2-a is a drawing of $\frac{1}{4}$ -inch open-tip mono jack, and Fig. 2-b is its schematic symbol. Figure 3-a is a drawing of a $\frac{1}{4}$ -inch open-tip stereo phono jack, and Fig. 2-b is its

schematic symbol. There are also other jack configurations for closed tips, closed tips and open rings, transfer, and "break" circuits.

Phono or phonograph plugs are primarily found in "line-level" audio or video systems. The names phono and phone are similar so they should not be confused. Today phono plugs are referred to as RCA plugs. Figure 4 shows one version of the RCA phono plug. It has a 0.125-inch diameter or 3.1 mm tip. RCA connectors typically connect a CD player to a stereo receiver with a shielded cable.

The XLR connector

The plugs previously described are single-cable connectors. Audio components can have a large number of jacks, so a large number of cables might be required. Consequently, the space behind a rack housing those components can look like a maze. In addition to being unsightly, this situation can make it difficult to locate and trace individual cables. Moreover, it can lead to the making of incorrect connections.

One solution is the use cables terminated with XLR connector plugs. Sometimes called Cannon plugs (a reference to the manufacturer, now ITT Cannon), they are standard input and output connectors for professional and some high-end (semiprofessional) audio equipment.

Figure 5-a shows a three-contact keyed XLR in-line socket, and Fig. 5b shows a three-contact keyed XLR in-line plug. It is possible to insert the plug in its jack in only one way. There are also five-contact keyed XLR connectors.

XLR connectors typically connect low-level signals from microphones to mixers or public address systems. They are also used to connect linelevel signals such as the outputs and inputs from mixers and signal-processing circuits to power amplifiers.

XLR connectors are balanced connectors, which helps to reduce any



FIG. 3— PHONE JACK: 1/4-inch stereo jack (a) and its schematic symbol (b).

noise and hum pickup by the system in which they are installed. Figure 6-a shows an XLR panel socket, and Fig. 6-b shows an XLR panel plug.

XLR adapter/transformers can adapt a microphone with a $\frac{1}{4}$ -inch plug to a public address amplifier or mixer that has an XLR input jack. There are also adapter/transformers that will adapt a professional microphone with an XLR plug to an amplifier or mixer with a $\frac{1}{4}$ -inch input jack.

Internationally accepted standards as well as the American EIA standards



FIG. 4—RCA PLUG is formed from sheet brass, and the tip and sleeve are insulated by phenolic.

RS-221 (10-79) and RS-297A (6-70) specify the pin numbers on XLR contacts. On a three-contact XLR, pin 1 is the ground connection, pin 2 is the audio (in-phase) connection, and pin 3 is the audio (return or out-of phase) connection. In Fig. 5-a, contact 1 is at the left, contact 2 is at the right and contact 3 is at the bottom.

On five-contact XLRs, the first three pins have the same functions as those on the three contact version with pins 2 and 3 assigned to channel 1. Pins 4 and 5 for channel 2 have the same functions as pins 2 and 3, respectively.

When the male and female connectors are joined together, pin 1 of the male makes contact with pin 1 of the female, establishing the ground connection before pins 2 and 3 are connected simultaneously. XLR connectors are usually used in balanced circuits with a ground connection, but they can function in unbalanced circuits.

There is no standardization as to 136 the way the wires of the cable are



FIG. 5-IN-LINE XLR CONNECTORS: jack (a) and plug (b).

connected to the terminal points of the plug or jack (also called a socket).

Unbalanced connectors

Unbalanced connectors for audio equipment were first introduced in the 1930's, but were modified for military equipment during World War II. One example is the PJ-042 military-style connector. After the war, a $1/_4$ -inch phone jack, based on the PJ-042, became popular for consumer audio applications. The original military design was simplified to make it easier to manufacture. the RCA connector.

The RCA connector was and is made by forming flat brass stock into a tulip shaped sleeve and a cylindrical tip insulated by phenolic insulation. This connector could be made in high volume and was later adopted as the connector of choice for the Hi-Fi industry that was forming in the 1950's. However, the $\frac{1}{4}$ -inch phone jack still offers lower contact resistance, which means less noise introduced into the system. It also offers better strain relief for portable cords.

In the mid 1950's, while stereo was



FIG. 6-XLR PANEL CONNECTORS: jack (a) and plug (b).

One of the early manufacturers of the 1/4-inch phone jack was Switchcraft, Chicago, Illinois. Although the Switchcraft products were successful, a lower cost audio connector developed by Cinch Connectors of Elk Grove Village, Illinois, became even more popular because of its lower cost. It has since become known as still in its infancy, a need developed for two conductors with a shield to carry the left and right signals and their common ground. The PJ-068 military standard connector, similar to the $\frac{1}{4}$ -inch jack but with three contacts, had been developed by then.

With the PJ-068 as its model, a Continued on page 159

SCHEMATIC SOFTWARE

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in that they can draw schematics, but they don't create netlists, generate a bill of materials (BOM), or lay out PC boards. Their sole purpose is to put your circuit designs on paper using schematic symbols. Most drawing programs can create schematic diagrams, even Windows' Paintbrushbut it takes a lot of time to draw them. Schematic draw programs simplify the process by "canning" common symbols, such as resistors and transistors, in a library file, where they can be called on screen, placed quickly, and wired together. In this section we look at three schematic draw programs ranging in price from \$20 to \$99.

Product:	Schematic Plus
Version:	0.4
Price	\$20
Platform:	Windows
Publisher:	Steven Poulsen

For rock-bottom price you can't beat Schematic Plus, a Windows-based shareware program written by Steven Poulsen. Despite its low \$20 asking price, Schematic Plus has some advanced editing features such as automatic annotation. It doesn't generate netlists or have circuit-simulation capability, but the program is easy to use and certainly powerful enough to draw schematics if you can live without zoom and block functions. Interestingly, the program saves the drawings to files in the Windows' MetaFile format (WMF), which means you can import your schematics directly into text and graphics applications, such as Word for Windows. That's the only way you can print the schematic to hard copy; the program itself doesn't have any kind of printer support.

ÂÎl wiring is orthogonal (drawn only horizontal or vertical). Devices are placed on the drawing one at a time by highlighting its name from a component library. Unfortunately, re-



FIG. 3—WHILE AUTOSKEM IS DIFFICULT TO USE, it will run on any PC which has just 384K of RAM and a single floppy drive. Which is perfect for owners of an 8088 or 286 system.



FIG. 4—TURBOCAD SCREEN. The nice thing about using a CAD program to draw schematics is that the software is a lot more versatile.

petitive part placement isn't supported. The component library contains about a dozen or so devices, some of which you'd seldom use, like the node summer (see Fig. 2). So be prepared to spend a fair amount of time creating device symbols with the library editor. However, you don't have complete freedom when designing new devices because library symbols are built on a 68-by-68 grid, and a dot takes up nine grid squares. That limits you to about 15 pins per device. So if you need to create a 68-pin IC, forget it. According to the author, the library editor will improve considerably as the software advances from version 0.4 to 0.6 over the next few months. Library devices are saved as individual files in a directory, and you can sort parts by storing them in different subdirectories.

Overall, Schematic Plus is an 137



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FIG. 5—ELECTRONICS PARTS-BIN is freeware created by Jim Marinelli that lets you build schematics from clip art.

excellent program for drawing and importing schematics into documents. And it's a dynamite buy for the money. The program was originally available to American Online subscribers only, where you can find Steven Poulsen should you need technical support, but it's now available on the Gernsback BBS (516-293-2283, v.32, v.42bis), CompuServe, and Internet under the filename SCH06.ZIP. Download time is about eight minutes.

Product:	AutoSkem
Version:	1.41
Price	\$29
Platform:	DOS
Distributor:	BSOFT Software

If you're a starving student saddled with an old circa-1986 PC, you might want to consider AutoSkem—a very inexpensive (\$29) DOS-based schematic drawing program that lets you create simple schematics on vintage computers that don't have the resources to run Windows. All it takes is any IBM-PC or PC clone, including a lowly 8088-based machines with 384 kilobytes of RAM and a floppy drive. Be warned, though, Auto-Skem isn't an easy program to use, and it lacks some basic editing features.

While not nearly as sophisticated or powerful, AutoSkem has the look and feel of Schema, a very-successful, high-end schematic-capture program from the 80's that was just recently put out to pasture. Drawing and editing commands are located in a menu column displayed along the right side of the screen (see Fig. 3), and are selected by the cursor keys or mouse. A handful of the more basic commands, like grid on/ off and cross-hair sizing, are controlled by the keyboard's function keys, and are not listed in the menu.

Schematics are constructed from three basic objects: lines. symbols, and text. Lines aren't orthogonally locked, making it easy to draw diagonals but harder to square corners. Symbols representing electronic devices are stored in symbol libraries, each of which can hold up to 100 symbols. While there's no limit to the number of libraries you can have, a schematic can be drawn using components from only a single library. If the component doesn't exist in the selected library, it's simply not shown in the drawing.

Part designations and values are placed on the drawing with a text editor, and are not a part of the device itself. Both symbols

and text can be rotated and sized before placement-but not after. For devices that need more than 30 lines to construct (the maximum you can use for a symbol), AutoSkem offers an object editor, which is basically a block capture feature that saves the drawing to an object file.

Like symbols, objects can be called from a menu and placed anywhere in a drawing. Their advantage is that they can contain text and symbols, as well as lines. Their disadvantage is that they can't be scaled or rotated.

The drawing area is 1000 by 1000 units square, but only one-ninth of it is visible on the screen at any time, making it difficult to draw large schematics. Yes, you can zoom out to see the overall picture, locate the area of interest, and zoom back in. But you can't run commands from the zoom-out mode.

So a lot of initial part placement must be done by guess and by golly. Compounding the problem is that devices can't be edited after placement; they can only be moved. And if the object to be moved is bigger than the viewing window, it has to be erased and placed again-a tedious job, indeed.

AutoSkem's greatest attribute is undoubtedly its print features, which, for a DOS drawing program, are quite remarkable. Drawings can be output to a dot-matrix printer, LaserJet printer, or HPGL plotter. All or any part of the drawing can be selected for printing, and the selected area of the drawing can be rotated and sized to fit the paper.

Six years ago AutoSkem would have been a big hit at any price. But today it's as dated as the dinosaurs it's designed to run on. However, \$29 is a tempting carrot. So if you'd like to try this program, you can find an abridged, but fully functional version of AutoSkem on the Gernsback BBS or America Online under the name AUTO-SK14.ZIP. Download time for this program is about four minutes.

Any drawing program can draw schematics. The drawback here is that creating the symbols (transistors, inductors, ICs, and so on) takes a lot of patience and time. However, there are some CAD drawing programs that have electronic symbol libraries—like TurboCAD for Windows. Unlike some drawing programs with similar libraries, such as Auto-CAD, TurboCAD's \$99 price won't break the piggy bank. There's even a library module with PC board layout symbols (see Fig. 4).

The nice thing about using a CAD program to draw schematics is that it has a lot more to offer, like unlimited zoom, multiple drawing layers, and extremely large drawing sheets. Although you can find schematic capture programs with the same features, they are generally very expensive.

The basic TurboCAD package comes with a limited number of schematic symbols that include a capacitor, resistor, and a handful of TTL ICs. For real work, though, you'll want to buy one or more of the 20 extracost libraries IMSI sells for use with TurboCAD. But as you add libraries to the setup, cost goes up. Fortunately, the libraries aren't that expensive. For example, the TTL, ECL, and CMOS libraries sell for \$29.95 each. Or you can get all three, along with several other electronic modules, for just \$79.95 when you purchase the Electronic Design library.

A shortcoming in using a CAD program to draw schematics is that it can be overkill some of the time. You have a lot more power than you need, so some things are a little more

Product:	Electronics Parts-bin
Price	Free
Platform:	Windows
Publisher:	Jim Marinelli



FIG. 6-BOARDMAKER's screen could not be captured, which is why you see a printout instead.

complicated than they should be. And the learning curve is steeper. That's why TurboCAD is not recommended for hobbyists who only need to draw simple schematics. But if you need a drawing program to create complex schematics frequently, then TurboCAD could be just the ticket you're looking for.

Electronics Parts-bin lets you build schematics from clip art (see Fig. 5). The clip art library of general purpose electronics components is contained in a file called PARTBIN1.PCX that's available on the Gernsback BBS, and can also be found in the software forum of America Online. Download time is about 11 minutes. To use the library, open up PARTBIN1.PCX in a drawing program like Windows' Paintbrush and copy the component you wish to the clipboard. Open a new or existing file and paste the component in place. Lines and text are provided by the drawing program.

Schematics with PCB layout

One of the joys in life is taking an idea from concept to a working device. The project doesn't have to be complicated or earth shaking—but it surely has to end in an etched printed-circuit board. If the design is simple consisting of a couple of ICs and a fistful of passive parts—it's easier to do the PC board layout manually rather than run the design through a bunch of netlists. For such manual design, the software should be simple and to the point. The following three programs—all DOS-based British imports fill the bill. They let you draw schematics plus manually layout a PC board using screens and commands that are common to both.

BoardMaker I 1.5 \$95 DOS Ohio Automation
Uhio Automation

If you're in the market for an economical schematic drawing program with PC board layout that's easy to use, yet powerful enough to handle large designs, you'll fall in love with BoardMaker I. This program can lay out circuit boards up to 17 inches square. It has advanced drawing and editing features normally found in programs costing several hundred dollars, and it supports a wide range of printers, pen plotters, and photoplotters. You can buy BoardMaker I for just \$95.

Basically, BoardMaker I is a DOS-based PC board layout pro-

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	AutoSkem	BoardMaker	CADPAK II ISIS Supersketch	CircuitMaker	Easv-PC	Easv-PC Prn	Electronics Workhench	EZ-Route Std F7-Lonic	Schematic Plus	SunarCAD	SuperCAD for Windows	TurboCAD for Windows	WinCebama
Price	\$ 29		\$ 159		\$195	\$349	\$299	\$249	\$20	899	\$149	\$99	\$179
Format	DOS	DOS	DOS	Windows	DOS	DOS	Win/DOS	Windows	Windows	DOS	Windows	WIndows	WIndows
RAM required	384K	512K	640K	2MB	512K	640K	4MB	512K	2MB	640K	1MB	2MB	2MB
Hard disk required	z	z	2MB	7MB	z	5MB	4MB	3MB	z	Z	2MB		10MB
Schematic	٨	۲	7	>	٨	>	۲	٢	۲	٢	Y	٨	۲
Capture	z	z	z	7	z	٨	7	۲	z	7	٢	Z	Y
PCB netlist	z	Z	z	z	z	٨	Y(1)	7	z	۲	٢	z	Z
Simulation netlist	z	z	z	7	z	7	Y(1)	۲	Z	Y	Y	z	۲
Design rule check	z	z	z	z	z	>	۲	Y	Z	Y	٨	z	٨
PCB Layout	z	٢	٨	z	٨	>	Z	٨	z	N	Z	z	Z
Circuit Simulation	z	z	z	~	Z	۲	٢	Z	Z	٨	٨	z	z
Analog	Z	z	N	Y	N	Y(1)	٨	Z	z	Y(1)	Y(1)	z	Z
Digital	Z	Z	N	Y	z	Y(1)	۲	z	z	Y(1)	Y(1)	z	z
Drawing Features													
Ortho disable	n/a	٢	Y	z	٢	7	z	7	z	Y	٢	Y	z
Diagonal wires	Y	٢	٨	z	٢	>	Z	٨	z	Υ	٨	λ	z
Auto annotate	N	N	Y	Y	z	7	Z	7	7	٢	٢	z	Y
Supports buses	N	N	Y	Ý	z	Z	z	7	z	٢	٨	z	7
Repeat object placement	N	Υ	٢	Y	٢	7	z	7	z	Υ	7	٢	z
Rotate part prior to place	۲	٢	٢	Y	Y	Y	Z	N	٢	M	¥	٨	٨
Zoom range/on-the-fly	n/a	٢	٢	20:01	٢	۲	Z	۲	n/a	draw to fit	24:1	n/a	10,000:1
Auto pan	z	7	۲	z	٢	Y	Y	Y	Z	Z	z	۲	z
Find/search	z	z	٨	z	z	z	Z	N	Z	Y	٢	z	z
Editing Features—Placement													
Rotate	z	۲	۲	*	۲	۲	~	7	7	Y	γ	Y	Y
Mirror	z	٢	۲	7	۲	7	z	7	z	٢	۲	Y	٢
Find	z	z	۲	z	z	z	z	z	Z	Υ	٢	N	z
Undo	z	z	۲	Y(2)	z	z	Z	7	z	٢	٢	Y	Y
Rubberband	z	٨	۲	7	٢	٨	7	۲	z	7	7	Z	Y
Block	-												
Move	> :	>	> ;	> :	>	> :	7	7	z	٢	7	*	7
HIDE	z;	z	z	z	z	2 :	z	z	z	z	z	>	z
Delete	× :	× :	-	> :	~	> :	>	> :	>	٨	>	٨	٢
Save	7	~	>	~	>	>	>	>	z	z	z	٨	7
Part reterence	;		;	;	:	;		:		-			
Wide		6/n	> >	> 2	2 2	> 2	zz	zz	z 2	z>	z>	n/a	>
Delete	2 >	e/u	- >	2 >	2 2	2 2	2 2	2 2	2 2	- 12	- 2	B/H	- >
Adjustable font size		n/a	. 2	z	2	: z	2	2 2	2	2 2	zz	n/a	- 2
Global reannotate	Z	n/a	Υ	Y	z	z	z	z	z		. >	n/a	Z
Library													
Screen directory	۲	۲	٢	٢	z	Z	Υ	Y	Y	Y	۲	7	۲
DeMorgan	z	z	۲	z	z	z	z	z	Z	Y	Υ	z	٢
Printer Output	;			:		-							
FIT TO page	>	Y(3)	2	z	z	z	>	n/a	n/a	z	n/a	n/a	n/a
	n/a	E/U	n/a	n/a	n/a	n/a	n/a	> :	n/a	> :	>	n/a	n/a
	n/a	B/1	n/a	> ;	n/a	e/u	n/a	> :	n/a	>	>	n/a	n/a
Software control	7	>	~	~	>	>	>	>	n/a	>	>	n/a	n/a
Flotter Output	>	>	>	2	-	-		-1-	-1-				
Hit to page Maximum nane size /inches/	۲ ۵ کار	Y 17~17	۲ <u>۲ مم م</u> ر	N × 24	N 17~17	N 17~17	n/a	n/a	n/a	6/u	n/a	n/a	n/a
Software control	A × v			N N	N X N	11 × 11	D/a	40 X 40	n/a	p/u	94 × 94	b/1	n/a
	1) Extra cost		2) Limited (Limited to move and change only	ance only		3) Except dot-matrix printers	latrix printers		5		1	4

TABLE 1

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gram with access to a library of schematic symbols. Although the two programs share the same drawing editor, they don't share libraries or communicate between each other. In other words, you don't have access to the schematic library when it is in the PC board layout mode, and you can't import schematics into the PC board program for layout. Each program is independent of the other, and you can run only one at a time. An opening menu gives you the choice between schematic draw and PC board layout. An advantage of sharing the same drawing editor is that the commands are identical, which means you don't have to adjust to a new set of commands as you switch from one program to the other. The commands are easily accessed from seven pull-down menus, plus there's a keyboard shortcut for every command or command string.

BoardMaker I's schematic library comes with a total of 75 parts that include a good spattering of analog, logic, and discrete components, as well as a handful of lesser used devices like a microphone and Schottky diode. But as you've probably guessed, a library this small and diverse doesn't contain all the devices needed for the example schematic. That's why two ICs had to be created from the library editor. Like the schematic and PC-board layout editors, the library editor is a stand-alone program that's listed as one of the opening menu choices. And like all BoardMaker I programs, the library editor has the same screen setup and command menus, which makes the BoardMaker I package really easy to learn and use. New devices can be made from scratch or by modifying an existing component; there is little difference between the two. Each library can hold up to 250 symbols, depending on the symbol's complexity, and there's no limit to the number of libraries you can create.

The components are easily accessed from a pull-down menu. Parts can be rotated and mir-



FIG. 7—EASY-PC IS BASICALLY A PC board layout program with a schematic component library, and it is the precursor of BoardMaker I.

rored before and after placement. The ability to zoom in and out and pan during placement and editing are two very powerful features that make working with this program a real pleasure. Repeat placement is provided, and there's automatic backup that saves your drawing to file at scheduled intervals without you having to think about it. Output can be sent to dot-matrix, LaserJet, and PostScript printers, plus HPGL and DMPL penplotters. If you don't mind fiddling with sizing parameters, the drawing can be made to fit the page for everything but a dot-matrix printer. Gerber and NC-Drill files-used by professional circuit-board manufacturers-are also supported.

Unfortunately, BoardMaker I wouldn't allow a screen shot to be made of its screen, which is why you see a printout in Fig. 6 instead. So you don't get to see the pull-down menus that include the component library. For that drawing, two ICs had to be created. The boxy European resistor symbols could easily be redrawn to U.S. standards.

BoardMaker I is extremely easy to use, and it is full of features you wouldn't expect in a program costing so little. And, unlike many schematic draw/ PC board layout packages, BoardMaker I schematics are so good looking that you won't be embarrassed to publish them. Good looking schematics, the ability to lay out large circuit boards, and low cost just might make this the buy of the century.

Product: Version: Price Platform:	Easy-PC 10142 \$195 DOS
Platform:	DOS
Distributor:	BSOFT Software

When the author first started testing Easy-PC, he thought he had picked up an old copy of BoardMaker I and loaded it by mistake. Everything that Easy-PC does—from the opening menu, to the drawing/editing tools, to output support smacks of an early version of BoardMaker I. And for good reason: both are written by the same person. Essentially, Easy-PC is the forerunner to BoardMaker I, and while the two are very much alike, there are some significant differences. For openers, Easy-PC lists for \$195, which is \$50 more than BoardMaker I.

Like BoardMaker I, Easy-PC is a DOS-based PC board layout program with a library of schematic symbols in addition to



FIG. 8—ISIS SUPERSKETCH'S EXTENSIVE LIBRARY contains all the symbols needed to create the *Electronics Now* benchmark schematic.

component footprints and pads. The schematic draw and PC board layout programs are separate, and access to the programs is through a main menu. The Easy-PC programs share a common editor, and the commands are the same for all programs. Up to this point BoardMaker I and Easy-PC are virtually identical. The only difference is the look of the menus and the way they're organized. But here's where the two packages part ways.

While Easy PC's menus are easy enough to use, it takes time to navigate them because the commands are nested and one menu often spawns another. Fortunately, each menu has a keyboard equivalent that frequently reduces a lot of mousing around to a single keystroke (see Fig. 7). However, the keyboard commands aren't intuitive or easily remembered, which is why the keyboard reference chart was posted in a conspicuous spot.

You'll want to keep a list of the symbol libraries close at hand, too, because (unlike Board-Maker I) Easy-PC's screen lacks a library list. Parts are selected by typing their names from a command line—which forces you to memorize them or refer to the symbol charts found in the loose-leaf user's guide. But

after thumbing through these lists to the point where the pages looked like a well-read paperback novel, the author gave up and tacked the pages to the wall above the keyboard reference chart. Despite a high parts count, the schematic library lacks some very basic components, including a variable capacitor and an op-amp. That's because over 80% of the nearly 500 devices are dedicated to TTL logic. For the example drawing, symbols had to be created for two analog ICs plus a variable resistor. Thankfully that's easy enough with the library editor also found as an option in the opening menu. But again, you have to quit what you're doing to access it, because only one Easy-PC program can run at a time.

Printer and file output is good to excellent, depending on your application, with support for dot-matrix, LaserJet, and HPGL printers and plotters. While the drawing can't be sized to fit the page, it can be scaled in increments of $\frac{1}{4}$, $\frac{1}{2}$, 1, 2, and 4 times. Also included are utilities that generate Gerber files and standard NC Drill files, both of which are used by service bureaus to etch and drill finished printed-circuit boards.

Easy-PC is not BoardMaker I, and visa versa. BoardMaker I is



cheaper and easier to use. On the other hand, Easy-PC has a larger component library, and its files are upgradable to Easy-PC Pro, a fully-featured schematic capture and PC board layout program. The one you choose should depend on your present and future needs. For quickie projects, BoardMaker I is the best choice. For more ambitious designs go with Easy-PC because it has Gerber file and NC Drill support. But read on before deciding on either.

Product:	CADPAC II
Version:	2.28
Price	\$159
Platform:	DOS
Publisher:	R4 Systems Inc.
1	

CADPAC II is a combination of two DOS-based circuit design applications: ISIS Supersketch schematic draw and PC board II printed circuit board layout, both from Labcenter Electronics. Supersketch represents the entry-level program in the ISIS (Intelligent Schematic Input System) family of design software which includes ISIS Designer and Ares. While you'll find Supersketch's learning curve steeper than most schematic draw programs, you'll discover the advanced features, like auto wire routing and global device replacement, well worth the effort. And although the learning curve is steep, the price certainly isn't. At \$159, CADPAC II is just \$10 more than Easy-PC, which doesn't hold a candle to it.

Supersketch comes with several thousand drawing symbols stored in seven library modules, which are divided into such groups as CMOS, TTL, and memory devices. In addition there's an empty user module that can be used as a staging area to assemble new libraries from parts moved or copied from other libraries. New devices are created from the main drawing and can be saved to any library. Often it's possible to reduce the library part count by splitting the device into prefixes

Continued on page 159

OVER/UNDER VOLTAGE

continued from page 40

Therefore, monostable multivibrator IC2 provides a ½-second delay to keep RY1 turned off long enough for the voltage across C5 to return to normal. When pin 7 of IC1 goes high it triggers IC2, and the high output at pin 6 of IC2 causes the output of NOR gate IC3 go low which turns Q1 off and de-energizes RY1.

Switch S2 lets you quickly check the circuit's operation at any time. Pushing S2 raises the voltage at the junction of R2 and R5 and turns RY1 off. Reset switch S1 restores normal operation. Switch S1 turns the circuit on initially. An LM7805 voltage regulator (IC4) provides a constant 5 volt supply for the integrated circuits and the relay control current. No power switch was added to the circuit,

- All resistors are 1/4-watt, 5%, unless otherwise noted.
- R1-7500 ohms
- R2—3800 ohms, 1% metal film (see text)
- R3—412 ohms, 1% metal film (see text)
- R4-100 ohms, trimmer potentiometer
- R5—7385 ohms, 1% metal film (see text)
- R6, R8-10,000 ohms, 1% metal film
- R7—2692 ohms, 1% metal film (see text)
- R9, R10-4700 ohms
- R11-39,000 ohms
- R12-10,000 ohms
- Capacitors
- C1, C4-0.1 µF, ceramic disk
- C2-22 µF, 10 volts, tantalum electrolytic
- C3---1 μF, 25 volts, tantalum electrolytic
- C5—1000 µF, 35 volts, electrolytic Semiconductors
- IC1—ICL7665 over/under voltage detector (Maxim or equivalent)
- IC2-74121 monostable multivibrator
- IC3-74LS02 quad 2-input NOR gate
- IC4—LM7805 + 5 volt regulator BR1—DIP bridge rectifier, 1 am-



FIG. 3—A COMPUTER-TYPE POWER INPUT LINECORD was used with an input socket mounted on the back of the case.

as the device is intended to be on all the time. Note that the neon indicator lamps contain internal current-limiting resistors that aren't shown in the schematic.

PARTS LIST

pere, 50 volts (Marlin Jones #5225-BR or equivalent)

- Q1-2N3904 NPN transistor
- Other components F1—3-ampere fuse, 115 VAC
- S1, S2—pushbutton switch, normally open
- RY1—Solid state, normally open AC relay (240 volts, 10 amperes, see text)
- T1—Power transformer, 120 VAC primary, 12 VAC secondary (Marlin Jones # 6485-TR or equivalent)
- MOV1-MOV3—Metal oxide varistor, 130 VAC, 10 amperes
- NE1, NE2—neon indicator lamps with built-in current limiting resistors
- PL1—grounded IEC-type line cord with IEC-type input socket and built-in fuse-holder (optional, see text)
- SO1, SO2—grounded AC receptacle (only one is necessary, others are optional)
- Miscellaneous: Project case, circuit board, hardware, 20-gauge stranded hookup wire, solder.

Note: The following items are available from TDL Electronics, 5260 Cochise Trail, Las Cruces, NM 88012-9736, 505-382-8175, Fax 505-382-8810:

- Etched and drilled PC board—\$8.50 postpaid
- Kit of PC board and all board-mounted parts (including power transformer)— \$39.95 postpaid
- PC mount power transformers, solid-state relays, and input and output connectors are available from:

Marlin P. Jones and Assoc., Inc. PO Box 12685

Lake Park, FL 33403-0685 407-848-8236

All Electronics Corp. PO Box 567 Van Nuys, CA 91408-0567 800-826-5432

Herbach and Rademan Co. PO Box 122 Bristol, PA 19007-0122 800-848-8001

• The aluminum case (model MC-5A) is available from: Sescom, Inc. 2100 Ward Drive Henderson, NV 89015-4249 800-634-3457



FIG. 4—THE INSIDE OF THE COMPLETED PROTOTYPE. The MOVs are soldered to the input power connector; the fuse is built into this connector and is not visible in the photo.

Construction

The circuit layout is not critical. You can build the circuit on a PC board or it can be wired point-to-point. A foil pattern is provided so that you can make your own board or one can be purchased alone or as part of a kit from the source given in the Parts List. All the logic including the power transformer mounts on the small, singlesided PC board.

A parts-placement diagram is shown in Fig. 2. Install the parts as indicated. Only one jumper wire is required in the position marked "J." Sockets for the ICs are optional. Remember to observe polarity on electrolytic capacitors C2, C3, and C5. Use No. 20 AWG stranded wire for all connections to and from the PC board and between the input and output receptacles. (All parts within the dashed outline in Fig. 1 are board-mounted, and connections outside the dashed lines are hard wired.)

Because the author did not keep a complete stock of 1% metal-film resistors (and you probably don't either), he designed the PC board to accept a series pair of resistors for R2, R5, and R7. That should make it easier to find a pair of 1% resistors that will combine for the correct value. If you have 1% resistors of the exact values for R2, R5, or R7, install a jumper in one of the locations and the resistor in the other.

Install the socket for an IECtype detachable power input linecord on the back of the case (see Fig. 3). Fuse F1 is built into this socket, and the three MOV surge protectors are soldered directly to its lugs. You can use a regular linecord if you like, and eliminate SO1. Just remember to install the fuse separately if you eliminate the socket. Also be sure to use a polarized threeconductor line cord and remember to connect the green ground wire to the case-if a metal case is used-and to the ground terminals of the output receptacles. Mount the neon panel lamps that indicate input and output power on the front panel along with the test and reset buttons.

Figure 4 shows the inside of the completed prototype. The

The Protector circuit is designed so that the voltage at the junction of resistors R2 and R5 is set to be 2 volts . when the AC line voltage is 115 volts (see Fig. 1). That voltage is somewhat arbitrary, but it does need to be above the 1.3-volt switch point of the comparators and well below the unregulated DC voltage from the power supply. Proportionally, the voltage at that point will decrease to 1.65 volts when the line voltage drops to 95 volts, and it will increase" to 2.26 volts when the line reaches 130 volts. These values (1.65 and 2.26) correspond to the switch points that set the width of the window.

A value of 10K for R6 and R8 sets the current through R5-R6 and R7-R8 (the current into the comparator is negligible). At the 1.3-volt switch point, a current of 130 microamperes will flow (1.3V/10K). Pin 6 is the high-side comparator so the voltage drop across R5 will be 2.26 - 1.3 = 0.96 volt. At 130 microamperes, a resistance of 7385 ohms is needed. For the low-side comparator at pin 3, the corresponding voltage drop is 1.65 - 1.3 = 0.35 volt. So R7 must be 2692 ohms.

Values for R2, R3, and R4 were calculated by recognizing that the current flow in this branch should be large compared to the flow in R5-R6 and R7-R8. We'll let this current be 5 milliamperes, and knowing that the voltage at the junction of R2 and R5 is 2 volts, the R3-R4 series combination should be 400 ohms. Because the flow in the parallel combination of R5-R6 and R7-R8 must be considered, the 400 becomes 460 ohms. A 100-ohm trimmer in series with 412 ohms will work. With the power transformer used in the prototype there is a 19-volt drop across R2. With a 5milliampere current flow, R2 must be 3800 ohms. $\mathbf{\Omega}$

PC board is mounted to the bottom of an aluminum enclosure with four ³/₈-inch long threaded metal standoffs.

Calibration

Set the trimmer potentiometer R4 to mid-range. If you have a variable transformer (Variac) and an AC voltmeter, you can check the low voltage turn-off point. Adjust the variable transformer to 110 to 115 volts as indicated on the voltmeter. Push the reset button and slowly decrease the input voltage until the relay opens. (A lamp plugged into the Protector's output makes this easy to observe.) Adjust R4 so that the relay turns off (opens) with an input of about 95 volts. Ω

BRIDGE CIRCUITS

continued from page 42

0.3% on the 100-ohm and 1megohm ranges where the R1/ R2 ratios are 1/100 and 100/0, respectively. To be a practical instrument, the Fig. 3 circuit must include a sensitive nullbalance detector. Figure 4 is a schematic for a $\times 10$ DC differential amplifier for use with an external analog volt-ohmmeter to form such a detector. This circuit must have its own own independent 9-volt power supply. The LF351 is a low-cost BIFET operational amplifier available from Motorola, Texas Instruments, National Semiconductor, and others.



FIG. 4—A DC NULL POINT AMPLIFIER for use with an external digital multimeter or analog volt-ohmmeter

FIG. 5—THREE ALTERNATIVE VER-SIONS of Wheatstone bridges follow the same rules as the circuit in Fig 2.



FIG. 6—HIGH-SENSITIVITY, SIX RANGE Wheatstone DC resistance-measuring bridge.

The LF351 has a JFET input for low input offset voltage and BIFET technology provides wide bandwidth and fast slew rates with low bias currents as well as low input offset currents and supply currents. It is the functional equivalent of the LF353, and the LF347 is a dual version.

An external volt-ohmmeter can be set to its 2.5-volt DC range for low-sensitivity measurements, or to its 50-microrange version of this circuit, and Table 2 lists its benefits over the circuit in Fig. 3. Its null sensitivity (which is proportional to the R3/R2 ratio at balance) is very high on all ranges, and varies from 0.003% at R3's full scale balance value, to 0.03% at one-tenth of full scale.

By confining all measurements to the top nine-tenths of the R3 range, one can measure all resistance values in the 1-

ampere or 100-microampere range for high-sensitivity. In the 100-microampere range, the circuit must first be balanced by short-circuiting its input terminals together and trimming the multiturn 10-kilohm SET-BAL-ANCE control for a zero reading on the meter.

Wheatstone variations

The Wheatstone bridge circuit of Fig. 2 can be configured in three other ways without invalidating the basic balance equations, as shown in Figure 5. In each form R1/R2 are known as the bridge's ratio arms. Notice that the bridge's signal-source and detector terminals can be transposed without upsetting the ciruit's balance equations. This also holds true for the various versions of capacitance and inductance bridges. The most useful Wheatstone bridge variation is that shown in Fig. 5-a.

Figure 6 shows a modern six-

TABLE 2 RESPONSE OF 6-RANGE HIGH-SENSITIVITY BRIDGE

Switch S1 Range	Bridge Range	Resistor R1 Value (Ohms)	R1/R2 Ratio (Ohms)	Bridge Null Sensitivity (Nominal)
1	0-10	10	1/1000	Proportional
2	0-100	100	1/100	to value of R3
3	0-1k	1k	1/10	to 0.003% at
4	0-10k	10k	1/1	full scale on
5	0-100k	100k	10/1	all ranges
6	0-1MEG	1MEG	100/1	





FIG. 7—AC POWERED Wheatstone bridge with variable ratio-arm balancing and headphone-detection.

FIG. 9—WHEATSTONE BRIDGE for determining both capacitive and inductive reactances.



FIG. 8---FIVE-RANGE resistance bridge, with rotary panel potentiometer scale markings (a) and typical ratio scale markings (b).

ohm to 1-megohm range with excellent null sensitivity.

A Wheatstone bridge can be energized from either an AC or DC source without altering the fundamentals of its operation. Figure 7 is the circuit for an ACenergized Wheatstone bridge whose balance condition is maintained with an infinitelyvariable pair of "ratio" arms consisting of potentiometer R1 which provides the resistive values of R1 and R2.

The sum of the resistive values of R1 and R2 equals the resistive value of potentiometer. The balance sensitivity of this circuit is high enough to permit detection with headphones.

Figure 8 shows a five-range version of the Wheatstone bridge circuit of Fig. 7. It covers a resistive range of near-zero to near-infinite ohms with its highest precision between 10 ohms and 10 megohms. Rotary panel potentiometer R1's ratio equals one when its wiper is at midrange. The diagram shows expected typical scale graduations for a calibrating this control potentiometer. These must be manually marked as described later in this article.

To measure with the Fig. 8 circuit, connect the bridge to a 1kHz sinewave generator, insert the unknown resistor R_x, and adjust rotary switch S1 and rotary panel potentiometer R1 until a null can be detected by listening with the headphones. When that occurs, the value of resistor R_x equals the value of the resistor connected to S1 multiplied by the scale value on potentiometer R1. A balance can be obtained on any range, but for the highest precision, the balance should occur with an R1 scale reading between about 0.27 and 3.0.

To calibrate the scale of panel potentiometer R1, insert a 10kilohm, 1% resistor in the R_x position. Then index rotary switch S1 progressively through its 100-ohm, 1-kilohm, 10-kilohm, 100-kilohm, and 1megohm positions. Mark the scale at each sequential balance point as 0.01, 0.1, 1.0 (midscale), 10, and 100.

Repeat this procedure with



FIG. 10-LCR BRIDGE with headphone null detector.

 R_x values that are multiples or submultiples of 1.5, 2, 3, 4, 5, etc., until the scale is completely calibrated, as shown in Fig. 8-b.

Resolution and precision

There are four important quality characteristics of a measurement bridge: (1). measurement range, (2) balance-sensitivity, (3) resolution, and (4) precision. The term resolution refers to the accuracy with which the R_x value can be read from the bridge's controls. For example, in Figs. 3 and 6, R3 gives a resolution of about 1% of full-scale if it is a manually-calibrated, linear, potentiometer, or about 0.005% of full-scale if it includes a four-decade resistor box.

The resolution of the bridge in Fig. 8 varies from $\pm 1\%$ at a "1" ratio to $\pm 2\%$ at a 0.3 or 3.0 ratio, to $\pm 5\%$ at a 0.1 or 1.0 ratio.

The term precision refers to the basic accuracy of the bridge, assuming that it has perfect balance-sensitivity and resolution, and equals the sum of the R1/R2 ratio tolerance and the tolerance of the resistance standard R3. If the R1/R2 ratio is set strictly by precision resistors, the ratio's precision equals the

TABL	.E 3
STANDARDS VS.	BRIDGE VALUE

Standards Value		Bridge Range	
R3	100 Ω 100 k 1 MEG	10 Ω -1 k 1 k-100 k 100 k-10 MEG	
СЗ	100 pF 0.01 μF 1 μF	10 pF-0.001 μF 0.001 μF-0.1μF 0.1 μF-10 μF	
L3	1 mH 100 mH 10 H	100 uH-10 mH 10 mH-1 H 1 H-100 H	

sum of the R1 and R2 tolerances. However, there are other techniques that permit resistors to be matched so that ratio errors are reduced to only about $\pm 0.005\%$.

For example, if the high-resolution Wheatstone bridge in Fig. 6 is made with R1 and R2 as 1% resistors and R3 is a manually-calibrated control potentiometer, the circuit will have a basic precision of only 3%. However, if the values of R1 and R2 are correctly matched, the precision of the bridge increases to ± 1.005 %.

The circuit's precision can be increased to $\pm 0.105\%$ with a $\pm 0.1\%$ multidecade resistor box in the R3 position. Bear in mind that additional errors can creep in when you are measuring very low or very high resistance values. The most likely cause will be the resistance values of switch contacts and leads when measuring low resistive values, and leakages when measuring high values.

The overall quality of a bridge depends on its balance-sensitivity, resolution, and precision. Thus, the circuit in Fig. 6 offers excellent sensitivity and very good resolution and precision. Consequently, it can function as the basic circuit of either an inexpensive, simple bench test instrument or as a component in a precision laboratory instrument, depending on how it is built.

By contrast, the bridge in Fig. 8 has intrinsically poor resolution and precision. This means that it is suitable only for fast, imprecise measuments such as might be satisfactory for servicing equipment.

Capacitance and inductance

An AC-energized Wheatstone bridge can measure reactance, resistance, capacitance (C) and inductance (L). The circuit in Fig. 9 is a modification of the circuit in Fig.7 to measure C or L values by replacing R4 and R_x with comparable reactances. This will work if C_x or L_x are reasonably pure and have 1-kHz impedances greater than about 1 ohm and less than 10 megohms.

The difficulty in trying to measure inductance with this circuit is that accurate inductors (for use in the Z4 position) are hard to find, and inductive impedances are only 6.28 ohms per millihenry at 1 kHz.

The only difficulty in measuring capacitance is that the C_x value is proportional to the reciprocal of potentiometer R1 resistance scale markings. If the basic bridge is to measure both R and C, two calibrated sets of potentiometer R1 scales are needed. This drawback can be overcome by fitting R1 with a reversing switch. An example of this is shown as the multirange LCR bridge of Fig.10. Thus, only a single scale as



FIG. 11—ALTERNATIVES CIRCUITS for powering and detecting nulls in Wheatstone bridge circuits.

When that occurs, the C_x value equals 100 nanofarads. This 100-nanofarad standard can then be connected to the bridge and used to create a 1 microfarad standard.

Figure 11 shows two alternative block diagrams. Figure 11-a, the first option, is to power both circuits from the same supply but to isolate the oscillator by transformer-coupling its output to the bridge, as in Fig. 11-a. The second option, Fig. 11-b, is to power the oscillator from its own "floating" supply. This second option is more efficient.

Figure 12 is the schematic for a battery-powered bridge power source. It can provide either a 9volt DC output or an excellent 1kHz sinewave output with a peak-to-peak amplitude of 5



FIG. 12—POWER SOURCE for bridge circuits capable of 9-volt DC and 5-volt, 1-kHz outputs.

shown in Fig. 8 is required.

The general purpose LCR bridge with headphone detector shown in Fig. 10 is a versatile instrument. Switch S2 permits it to be used with either internal or external L, C, or R standards. The midscale value of each range is equal to the value of the standard assigned for that range, as listed in Table 3.

When this instrument is cali-

brated, it can be help to create its own alternative measurement standards. For example, if an accurate 10-nanofarad capacitance standard is inserted across the EXTERNMAL STANDARD terminals, a 100-nanofarad standard can be created by moving potentiometer R1 to the "10" position and then connecting capacitors in parallel across the "X" terminal for a null balance. volts. The oscillator is a diodestabilized Wien bridge oscillator that works from a split power supply derived from the battery through resistors R1 and R2.

The circuit has a low-impedance output, and consumes a less than 4 milliamperes of quiescent current. To set up the oscillator, connect its output to a suitable oscilloscope and trim potentiometer R1 to give a reasonably pure sinewave output of about 5 volts, peak-to-peak. Ω

MEMORY MODULES

continued from page 36

on a DRAM's capacitors will leak off, so it must be *refreshed* regularly. Refreshing data in a DRAM can be accomplished in several ways: each method calls for different combinations of RAS. CAS. and the address lines. Most modern computers have a special DRAM controller, in the form of an application specific IC (ASIC), that must manage both the memroy refresh function and data storage and retrieval chores.

DIP memory sources

Memory can be salvaged from

TABLE 1-DRAM PART NUMBERS AND CONFIGURATIONS

256K x 1 (16-pin DIP)	256K x 4 (20-pin DIP)	1Meg x 1 (18-pin DIP)	1Meg x 4 (20-pin DIP)
1256	GM4C256	GM71C1000	GM71C4400
41256	HM44256	HM511000	HM514400
1C256	M5M44256	M5M41000	M5M44400
2801	MCM514256	MCM511000	MCM514400
3C256	TC514256	TC511000	TC514400
6207	MT4C4256	MT4C1024	MT4C4001
62L07	KM44C256	KM41C1000	KM44C1000
MT1259	P21014	P21010	uPD424400
M5M4256	HM4514258	TM54C1024	MSM514400
TMS4256	uPD424256	AAA1M300	HYB514400

many sources, such as obsolete or inoperative computers, old memory boards, and EGA and VGA video adapters. If you try to salvage memory from an inoperative computer, exercise caution. It helps to know the history of the system that you are planning to cannibalize so that you do not inadvertently transport problems from one system to another. Memory is just as likely as any other component to have failed.



FIG. 2—THIS CONFIGURATION provides either 256 kilobytes or 1 megabyte of memory, populated in both cases with three DRAMs.

www.americanradiohistory.com

September 1995, Electronics Now



FIG. 3—SCHEMATIC of the 1-megabyte, nine-IC configuration.



SOLDER SIDE of the 1-meg board.

256KB, 3-IC PARTS LIST

IC1, IC2—256KB \times 4 DRAM IC3—256KB \times 1 DRAM IC4—not used C1–C3—0.1 μF , 50-volt, 0.3-inch center

1MB, 3-IC PARTS LIST

IC1, IC2—1MB \times 4 DRAM IC3—not used IC4—1MB \times 1 DRAM C1–C3—0.1 $\mu F,$ 50-volt, 0.3-inch center

256KB, 9-IC PARTS LIST

IC1–IC9–256KB \times 1 DRAM C1–C9–0.1 μ F, 50-volt, 0.3-inch center

1MB, 9-IC PARTS LIST

IC1–IC9–-1MB \times 1 DRAM C1–C9–-0.1 μ F, 50-volt, 0.3-inch center

ORDERING INFORMATION

The following items are available from AutoTime Corporation, 6605 SW MacAdam Avenue, Portland OR 97201. Voice: (503) 452-8577. Fax: (503) 452-8495, Faxback: (503) 452-0208. Internet: info@autotime.com. 256K/1M dual-purpose 3-IC DIP-to-SIMM conversion kit (DUAL3CDIP-EN, \$6.00), 256K 9-IC DIP-to-SIMM conversion kit (256K9CDIP-EN, \$6.00), 1MEG 9-IC DIP-to-SIMM conversion kit (1MEG9CDIP-EN, \$8.50). All kits include a PC board with through-hole plating, solder mask, silk screen, and decoupling capacitors. Memory not included. Add \$3.95 shipping and handling on all orders. AutoTime's trained memory technicians will help with memory identification, assembly problems, and explain other memory conversion products.

Fortunately, the memory is socketed in many older computer systems. If it is not, you must unsolder it. With a good desoldering system and some diligence, DIP memory can be extracted without damaging it. When desoldering, apply the least amount of heat for the shortest amount of time. Excessive heat will damage or destroy the semiconductor device.

If you must desolder, exercise caution. Safety is paramount. Wear safety glasses and a shirt with long-sleeves to protect against molten solder that can splash during device removal. After removing the ICs, clean all excess solder from the pins of the DIP with a soldering iron. Then straighten the pins with needle-nose pliers. Store the removed ICs in anti-ESD foam, and handle the devices follow-


COMPONENT SIDE of the 256-kilobyte board. SOLDER SIDE of the 256-kilobyte board.



FIG. 4—SCHEMATIC of the 256-kilobyte, nine-IC configuration.

ing proper ESD protection techniques.

Many memory components can be used with this project. Table 1 identifies many of the most common devices. If you're unsure about the identification or classification of a device, refer to a data book or contact the kit vendor listed in the Parts List.

Memory access time is an important consideration. As computer system clock speeds increase, DRAM access times must follow. Most of the newer systems require memory with access times of 70 to 80 nanoseconds; some even require 60nanosecond devices. If your computer's manual specifies the speed of the memory, you can safely use faster memory. But be wary of mixing different DRAMs with different speed ratings. The system might also have configuration jumpers that indicate memory access speed. Again, check your owner's manual.

Building the SIMM

The use of a PC board is mandatory for this project. Foil patterns are given here, but you can also purchase finished boards. To minimize potential clearance problems, remember that the layout of the components on each of the three boards is close. Use extra care during assembly, particularly if you make your own PC boards that do not have the solder mask that is applied to the commercial boards. Figure 5 includes three parts-placement diagrams, one for each of the three different board layouts.

Begin by tack-soldering the pins at the opposite corners of each IC. Before you complete the soldering, be sure that the component is seated flush against the board. Otherwise, adjacent boards might interfere. With tightly spaced motherboards, it might be necessary to trim the pins of the memory ICs after soldering them. If lead trimming is necessary, avoid cutting into the solder connections. That could fracture the solder and make the joint unreliable. After the board is fully as-



Please allow 6-8 weeks for delivery.







FIG. 5—PARTS LAYOUTS for the three-IC 256-kilobyte/1-megabyte board (a), the 1megabyte board (b), and the 256K board (c).

sembled, inspect your work carefully for solder bridges and cold solder joints. Also, check with an ohmmeter to be sure

Installation and test

If you plan to install your SIMMs in an IBM or compatible PC, install the 30-pin SIMMs in multiples of two or four. Both 386SX and 486SX computers require pairs of boards for memory expansion because their system data buses are 16 bits wide. All 386DX and 486DX systems require four boards for their 32-bit data paths. Some motherboards might require CMOS configuration changes

so that the hardware is aware of the new memory. Most newer PCs automatically detect new memory on power-up and then instruct you to run the setup program to "officially" configure the CMOS memory setup.

Some computers offer the option of complete or partial memory checks. After adding any new memory board, follow the complete checking scheme for several weeks.

Once your system is up and running, bring up your spreadsheet program and calculate your savings. Then, you can pat yourself on the back for recycling old memory. Ω

_____R

CO DETECTOR

continued from page 38

changes into account. The microcontroller also runs a lowbattery routine that will indicate when the battery must be replaced.

Another reason for the microcontroller in this CO detector is that the optical reflector needs approximately 20 milliamperes of current to operate, which would preclude battery power. To solve this problem, the circuit includes an LT1121CZ-5 five-volt regulator (IC3) that features a shutdown pin. The microcontroller turns the voltage

PARTS LIST

All resistors are ¼-watt, 5%. R1, R10, R11-510.000 ohms R2-150 ohms R3-680 ohms R4-2200 ohms R5, R12-20,000 ohms R6-1000 ohms R7-R9-10,000 ohms Capacitors C1-0.1µF, polyester film C2-100 pF, polyester film C3-0.33µF, polyester film C4-0.33µ.F, polyester film Semiconductors * IC1-LM358 op-amp IC3-LT1121CN8-5 5-volt regulator with shutdown feature IC2-PIC16C71 microcontroller IC4-LT1121CZ-5 5-volt regulator MOD1-Omron EE-SY148 optical sen-SOF Q1-2N3906 PNP transistor Q2-MPSA14 NPN transistor D1-1N914 diode BZ1- 6- to 12-volt piezoelectric buzzer Miscellaneous: 21/2- × 1- × 11/2-inch enclosure (Serpac No. 211), four 34-2 inch long No. 4-40 aluminum standoffs, 5-inch long 1/6-inch brass channel, 3-inch long 14-inch square aluminum, four rubber feet, PC board. solder Note: The following items are available from Cintron, Inc., P.O. Box 15, Roseville, MI 48066-0015: Preprogrammed PIC 16C71 microcontroller-\$24.95 Software on 5.25 or 3.5 inch disk-\$12.95 PC board—\$12.95 14.44 A kit consisting of preprogrammed PIC 16C71, software, PC board, optical reflector, and enclosure \$49.95 Check or money order accepted.* Please add \$3.00 to each order forshipping and handling. (Software and programmed microcontroller

provides for single use license.)

wie.

FIG. 4—PARTS-PLACEMENT DIAGRAM. The optical sensor (MOD1) and the 9-volt battery are mounted off-board.



CO DETECTOR FOIL PATTERN.

regulator on only long enough for the reflectivity of the CO reagent to be measured. That greatly extends the battery life allowing it to last approximately four months.

Another voltage regulator (IC4) supplies a constant 5 volts (V_{DD}) to the microcontroller. An LM258 op-amp connected as a voltage follower buffers the microcontroller's A/D channels. An MPSA14 Darlington transistor (Q1) turns buzzer BZ1 on and off. Resistor R5, diode D1, and capacitor C1 form a power-up reset circuit for the microcontroller.

Construction

A Dremel power tool will be handy for drilling and grinding when building the CO detector. The instructions provided here assume that you will use the enclosure specified in the Parts List. If you use a different enclosure, the construction details will vary. Just follow the same critical specifications for the mounting height and position of the optical detector and the dimensions of the card guide.

The first step is to assemble the printed circuit board and solder the components as



FIG. 5—AN ALUMINUM BAR holds the optical sensor in place. Brass U-channels are glued into slots cut in four metal standoff. The U-channels form a card guide.

shown in Fig. 4. Next place the 1/4-inch square by 1.1-inch long aluminum bar over the locator holes on the PC board, as shown in Fig. 5. Mark the locations for the two holes on the aluminum bar and drill and tap them for No. 4-40 screws; screws will pass through these holes to hold the aluminum bar on the PC board. Next, rotate the aluminum bar 90° and measure and mark a point that is centered about the length and width of the bar. Drill and tap the marked point for a No. 2-56 screw that will hold the optical sensor in place and allow for height adjustment above the reagent surface.

Mount the optical sensor on the aluminum bar so that the sensor end is facing down through cutout in the PC board



FIG. 6—THE END OF THE OPTICAL SENSOR should be facing down through cutout in the PC board. Remove the aluminum bar and optical sensor from the board and put it aside. The sensor must be 0.08 inch above the reagent surface.



FIG. 7—PLASTIC STANDOFFS molded to the bottom half of the case must be removed.



FIG. 8—CUT OR FILE a %-inch wide inch slot, 2½ inches long, ¾s of an inch from the bottom of the case.



FIG. 9—A $\frac{1}{100}$ -INCH WIDE SLOT, $\frac{1}{100}$ -inch deep, must be cut into each of the four $\frac{1}{100}$ -inch long aluminum standoffs, $\frac{5}{100}$ of an inch from the bottom end.



FIG. 10—PROTOTYPE CARD GUIDE/PC BOARD ASSEMBLY. This was assembled on perforated construction board so the component layout does not match the PC-board. The card guide would be the same design for either construction method.



FIG. 11—CUT OR DRILL SLOTS OR HOLES in the bottom half of the case so air will reach the sensor. Also place four $\frac{1}{2}$ -inch high rubber feet in the locations shown.

(see Fig. 6). Remove the aluminum bar and optical sensor from the board and put it aside. Grind down the plastic standoffs inside the bottom half of the case as shown in Fig. 7. Place the PC board inside the box, mark the four mounting locations on the bottom of the box, and drill a No. 4-40 hole at each location. Measure ⁵/₁₆ of an inch from the bottom of box and cut or file a ¹/₈-inch-wide inch slot 2¹/₂ inches long (see Fig. 8). The slot is where the CO detector card will be inserted.

Make the card guide by cutting a $\frac{1}{8}$ -inch-wide slot $\frac{7}{32}$ -inch deep into each of the four $\frac{3}{4}$ inch-long aluminum standoffs, 5/16 of an inch from the bottom end (see Fig. 9). Mount the four standoffs with the slots facing toward the center of the case with four 4-40 screws through the bottom of the case. Cut the ¹/₈-inch brass U-channel into two pieces, each 1.7 inches long. Epoxy one of U-channels into the standoff slots on one side (Fig. 5), and do the same for the other U-channel on the other side (the grooves in the channels should face each other). After the card guide is finished, mount the PC board to top of the standoffs with four 4/40 screws. Make sure the CO detector card can be inserted in and removed from the card guide; squeeze the middle of the brass channels slightly with pliers to make a snug fit.

Now remove the four screws from the bottom of the case and lift out the PC board with the card guide attached. Mount the aluminum bar and optical sensor assembly. Insert the CO card so that a feeler gauge can be inserted between the card and the optical detector. Set the gap between the sensor and the card to 0.080 inch, as shown in Fig. 6. The detector will not operate properly if the gap is incorrect.

After the gap is set, remove the sensor card. Figure 10 shows the completed prototype card guide/PC board assembly. The prototype was assembled on perforated construction board, so the component layout does not match the PC-board version, but the card guide would be the same for either. Solder wires from the optical sensor to the PC board, as shown in Fig. 4. (The sensor pins are labeled C, E, K, and Abe sure to connect the sensor pins to the board correctly.) Cut or drill slots or holes in the bottom half of the case to permit air to reach the sensor, as shown in Fig. 11, and place four ¹/₄-inch high rubber feet in the locations shown. Now put the PC board/ card guide assembly back into the case.

Software

You need a programmed PIC microcontroller to run the CO detector. The object code and executable file is posted on the Gernsback BBS (516-293-2283, v.32, v.42bis) as a file called CO-DECT. ZIP for those who have the necessary equipment to program their own PIC. However, preprogrammed PICs are available from the source given in the Parts List. A simplified flow chart is shown in Fig. 12.

Calibration and use

Before using the CO detector, it is important to keep in mind that it will only indicate the presence of carbon monoxide gas at the detector. The detector should always be located where CO will most likely show up first.

Additionally, the detector operates by "waking up" every two minutes and then checking the CO card for any change in reflectivity. So, any change in CO concentration will have a twominute lag.

The CO detector must be fitted with a new CO detector card. The cards are sold at most hardware stores. Write the date on the card when it is installed because the card must be replaced every three months. After dating the detector card, insert it into the slot with the yellow reagent side facing up until it hits the card stop.

Connect a fresh nine-volt battery to the detector. The beeper will sound twice when the battery is connected to indicate that the detector is working properly. If you remove the CO card, the buzzer will sound after a maximum two-minute lag. The sound will continue for 30 seconds and then repeat every two minutes or until a CO card is reinserted.





FIG. 12—SIMPLIFIED FLOW CHART for the PIC microcontroller software.

Slide the CO detector card back into the slot; the buzzer should stop sounding after a maximum of two cycles (each cycle is two minutes). The CO detector is now calibrated and ready to detect CO.

If, after initial power up, the buzzer should sound a shortduration beeping, there are four possible reasons for the alarm: One is that the carbon monoxide concentration is too high. This is a critical situation, and you should immediately look for the CO source. Two, the detector card is not inserted in the card slot properly. Three, the card has discolored. Discoloration can be caused by exposure to ammonia from cat litter boxes, ammonia, bleach, or sewer gas. If the detector card is

exposed to any of those gases, replace it. Four, the detector is in a high ambient light condition and the calibration routine has caused an error. If this occurs, repeat the power up routine.

If the detector's buzzer sounds once every two minutes, it is an indication of a low battery (less than seven volts). To test the carbon monoxide detector, just remove the CO card and the buzzer should sound within two minutes.

Warning: This detector has been used by the author for about a year and has worked well. It is important to note however, that the circuit has not been tested by any scientific laboratory for operation under all possible conditions. 0

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SCHEMATIC SOFTWARE

continued from page 142

and suffixes, as is the case with many TTL ICs. For example, 74LS161 and 74ALS161 are graphically the same. So rather than listing every permutation of the device, it saves time and space to create a list of prefixes which can be attached to any devices in a particular library.

Devices are placed on the drawing from an on-screen window that serves as both a bill of materials list and library menu for parts used in the current drawing. Device names are copied to this window from the various libraries, and then highlighted for placement on the schematic. The optional auto annotation feature can assign reference values to the devices as they are placed. Commonly used symbols, such as input/output terminals and ground, are not a part of the bill of materials, but are instead represented as icons along with the editing commands in the tool kit window (see Fig. 8). Symbols can be rotated in 90° increments and/or mirrored before and after placement. Repeated symbol replacement is supported, as is global find and replace—you can replace all 4001 gates with 4011 gates, for example. Moving device nomenclature is a simple drag and drop procedure.

To draw a wire, all you do is click once on its origin and once

AUDIO UPDATE

continued from page 136

new three-conductor commercial connector was developed to be compatible with the commercial/consumer 1/4 phone jack so that both the mono and stereo versions were compatible. Over the years, audio equipment has became smaller and more portable, as a result of transistorized circuitry, so the 1/8-inch or 3.5 mm connector was developed. Attempts to make the even smaller 2.5-mm version a global standard have not, so far, been successful.

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on its destination. Supersketch has an auto wire router that draws the line for you, automatically selecting the best route and whatever corners are needed to make it orthogonal. When moving devices, the autorouter automatically cleans up the rubberband lines into short, orthogonal lines.

Supersketch's output features are super slick, and include support for all the popular printers and plotters. Sizing isn't as flexible as in some programs, but the 4-to-1 scaling factor lets you print a 32-inch drawing on an $8-1/2 \times 11$ -inch page. A unique feature of Supersketch is its ability to produce output files suitable for graph-

DIN connectors

In your audio work you might encounter a DIN audio connector. The term DIN is the abbreviation for *Deutsch Industrie Normenausschuss*, originally a German standard, but now widely accepted throughout Europe. DIN cables usually have three or five wires.

The DIN connector has a keyway making correct connection certain while eliminating wiring errors. The DIN arrangement makes it impossible to transpose input and output connections, so polarity follows automatically. Nevertheless, with separate cables, be sure to distinguish between ics import into many popular word processor and desktop publishing programs.

When it comes to drawing schematics, Supersketch is unquestionably one of the best. The program is easy to use once you've scaled the learning mountain—and it abounds with advanced drawing and editing features that really help speed up your work. The best thing about the CADPAC II package, though, is its superb PC board layout program.

Next month, we'll continue with a look at schematic-capture packages including Super-CAD, WinScheme, Circuit-Maker, Electronics Workbench, EZ-Route, and Easy-PC Pro. Ω

the left and right channels when signal polarity is important.

One drawback to the DIN cable is that manufacturers assign different functions to the cable wires. This means that just because a DIN plug will fit a DIN jack, you can't be sure of the correct connection. Some components such as tape decks are supplied with one port for a DIN connector and other ports for RCA plugs.

Three- and five-pin DIN connectors are most common, but some have as many as eight pins. Because they all look alike, a three-pin DIN connector looks like an eight-pin DIN connector, so be sure to count the pins.

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BASIC Stamp I Module (BS1-IC) 8 general-purpose I/O lines 256-byte program space (100 instr.) 4-MHz clock (2400 baud serial, etc.) \$34, \$49 with optional carrier board



BASIC Stamp II Module (BS2-IC) 16 general-purpose I/O lines 2048-byte program space (600 instructions) 20-MHz clock (9600 baud serial, etc.) \$49, \$69 with optional carrier board

BASIC Stamps are perfect for many applications, from controlling model trains to monitoring factory sensors. They have 8 or 16 I/O lines, which can be used for a variety of digital and analog purposes. And to keep life simple, they're programmed in BASIC. Our special "PBASIC" language includes familiar instructions, such as GOTO, FOR...NEXT, and IF...THEN, as well as SBC instructions for serial I/O, pulse measurement, button debounce, etc.

The BASIC Stamp Programming Package contains everything you need to program Stamps using your PC. The package includes our editor software, programming cables, manuals, application notes, and free technical support. The package is available for \$99; Stamps and carrier boards must be purchased separately.

WHAT'S UP WITH THE STAMP II?

We've experienced countless delays with the Stamp II, and many callers have wondered what's wrong.

Well, there aren't any big bugs in the Stamp II, and the programmer didn't quit. When we placed our first ads last year, we honestly thought we'd be done in December. With every passing month, we thought it was just a month away, which is why we didn't pull the ads.

As this is written in mid-June, we feel confident that we'll be shipping beta units in 7-10 days, which will lead to real Stamp II units in July or August.

We're sorry for all the frustration we've caused, and we certainly don't wish to repeat the process anytime soon.

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controllers, we offer various development tools to meet your needs. If you're just getting started, you may be interested in our "Hobbyist Pack" PIC Programmer, which sells for just \$99. If you're more experienced with PICs, perhaps you could use the debugging features of our ClearView in-circuit emulators. We also offer assemblers, C compilers, prototyping boards, and data sheets. And for added convenience, we even offer the PIC chips themselves.

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	MODEL	Freq. (MHz) DC TO	GAIN (Typ. dB) At 100MHz	MAX. Power (@ 1dB Compr.) dBm	NF dB (Typ.)	Price \$ea. (Qty. 50)
MAR	MAR-1 MAR-2 MAR-3 MAR-4	1000 2000 2000 1000	18.5 12.5 12.5 8.3	1.5 4.5 10.0 12.5	5.5 6.5 6.0 6.5	.99 1.35 1.45 1.55
MAR SM	MAR-6 MAR-7 MAR-8	2000 2000 1000	20.0 13.5 32.5	2.0 5.5 12.5	3.0 5.0 3.3	1.29 1.75 1.70
Ж RAM	RAM-1 RAM-2 RAM-3 RAM-4	1000 2000 2000 1000	19.0 12.5 12.5 8.5	1.5 4.5 10.0 12.5	5.5 6.5 6.0 6.5	*6.40 *6.40 *6.40 *6.40
~	RAM-6 RAM-7 RAM-8	2000 2000 1000	20.0 13.5 32.5	2.0 5.5 12.5	2.8 4.5 3.0	*6.40 *6.40 *6.40
MAV	MAV-1 MAV-2 MAV-3 MAV-4	1000 1500 1500 1000	18.5 12.5 12.5 8.3	1.5 4 .5 10.0 11.5	5.5 6.5 6.0 7.0	1.10 1.40 1.50 1.60
MAV	MAV-5SM MAV-11	50-1500 10-1000	8.0 12.7	18.0 17.5	6.5 3.6	2.07 2.10
VAM	VAM-3 VAM-6 VAM-7 *Oty, 10	2000 2000 2000	11.5 19.5 13.0	9.0 2.0 5.5	6.0 3.0 5.0	1. 45 1.29 1.75
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