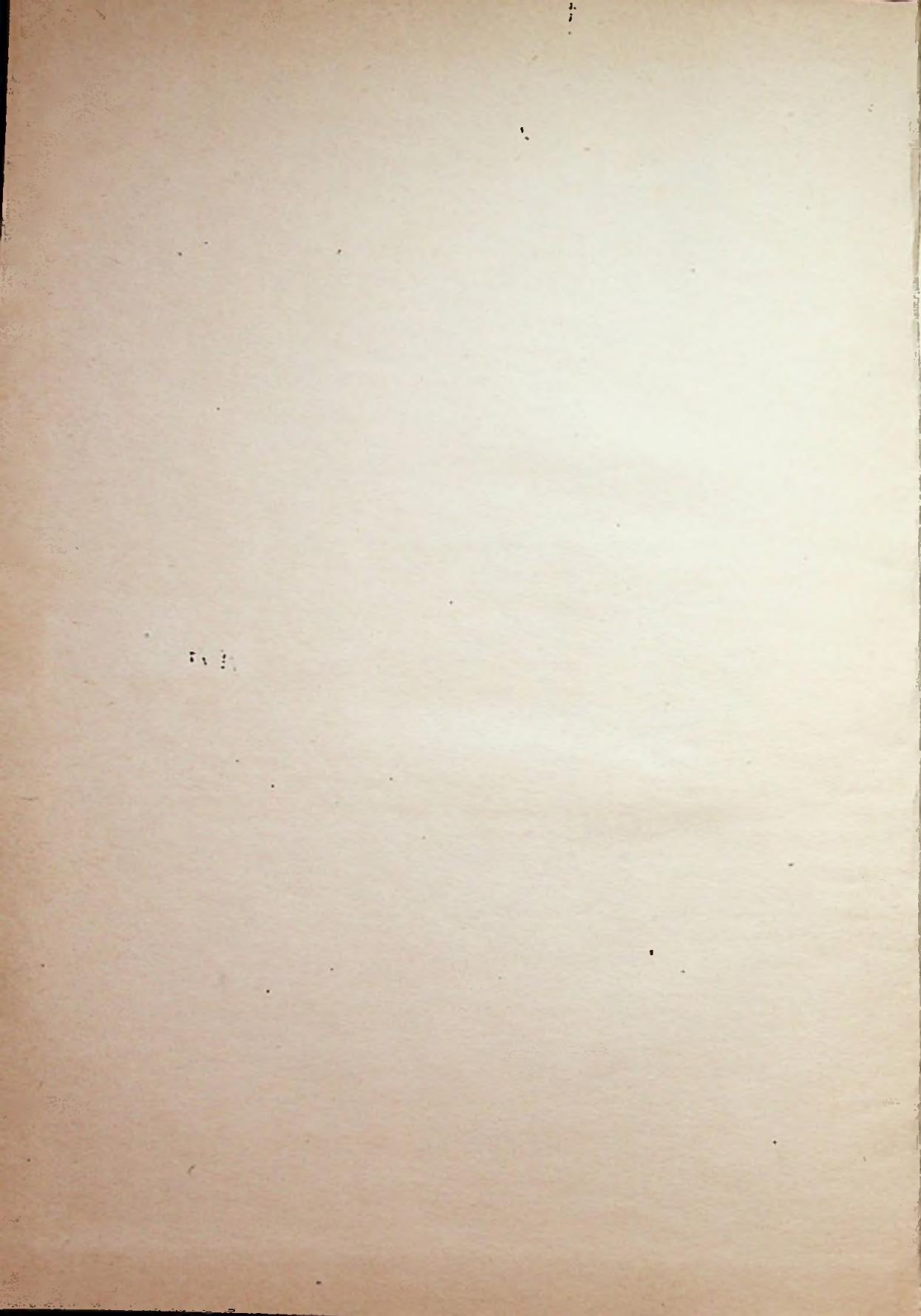




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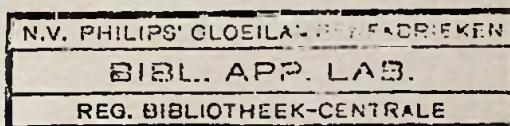
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RADIO CIRCUITS AND DATA

BY

C. A. QUARRINGTON
A.M.Brit.I.R.E.

A recognised Authority on Radio and Television



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FOREWORD

THE absence, among contemporary works on radio engineering, of a concise data book has become increasingly more apparent. It is true that numerous text-books contain a few tables, formulæ or other matter, but a shelf of books is necessary before all normal information of this class is available, and even then, much valuable time can be wasted in finding it ; when found it often leaves much to be desired, particularly from the point of view of covering modern needs or presenting facts and figures in a form that is immediately usable.

It is perhaps not unreasonable to form the conclusion that very little original work has been done on tabular data for more than a decade, old tables being published anew without being brought up to date or remoulded to suit changing needs.

The aim of this small volume has been twofold. Firstly, to present a very carefully selected series of circuits, data and formulæ ; and secondly, to reshape information so that it covers present-day requirements, is freed from errors and inconsistencies of the past, and in addition, includes both English and American standards.

The very complete index which follows these remarks permits quick reference to the information required, since the cross indexing reveals the presence of any particular subject under any reasonable heading. The main subjects are, however, briefly described below.

- (i) A representative selection of the circuits of modern receivers marketed by Britain's leading radio manufacturers is included primarily to illustrate present-day design tendencies.
- (ii) A selection of useful circuits designed to perform a variety of functions.
- (iii) A selection of relatively simple test gear circuits with full component values.
- (iv) Formulae and data selected and presented in such a manner that considerable saving of time can be effected by regular use. Particular care has been taken to so devise tables that all related data are available in a single table. The author wishes to thank Mr. S. S. D. Jones, M.A., of St. Peter's Hall, Oxford, who checked the formulæ and made a number of useful suggestions.
- (v) Valve equivalent tables and valve base connections.

This volume is essentially complementary to *Modern Practical Radio and Television*, since the latter is broad in its conception, whereas the former is precise in character and is intended for quick reference as opposed to systematic study.

C. A. QUARRINGTON.

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USEFUL CIRCUIT SECTION

THE following forty-one pages contain a selection of useful circuits presenting a broad picture of commercial practice and contemporary design and also show, by a series of "unit" circuits, everyday alternatives and refinements associated with specialised rather than domestic equipment. The commercial circuits in this section will also be found helpful when fault-finding, since the task of tracing connections without the assistance of the appropriate circuit diagram is simplified by reference to a selection of typical arrangements. A close study of these circuits will suggest ideas for the systematic stage-by-stage isolation of a faulty receiver, a procedure so essential to efficient servicing ; the use of the Information Bureau service for assistance when diagnosing a fault in a particular receiver is impracticable, since those who are able to select the appropriate facts and figures would not normally require such advice.

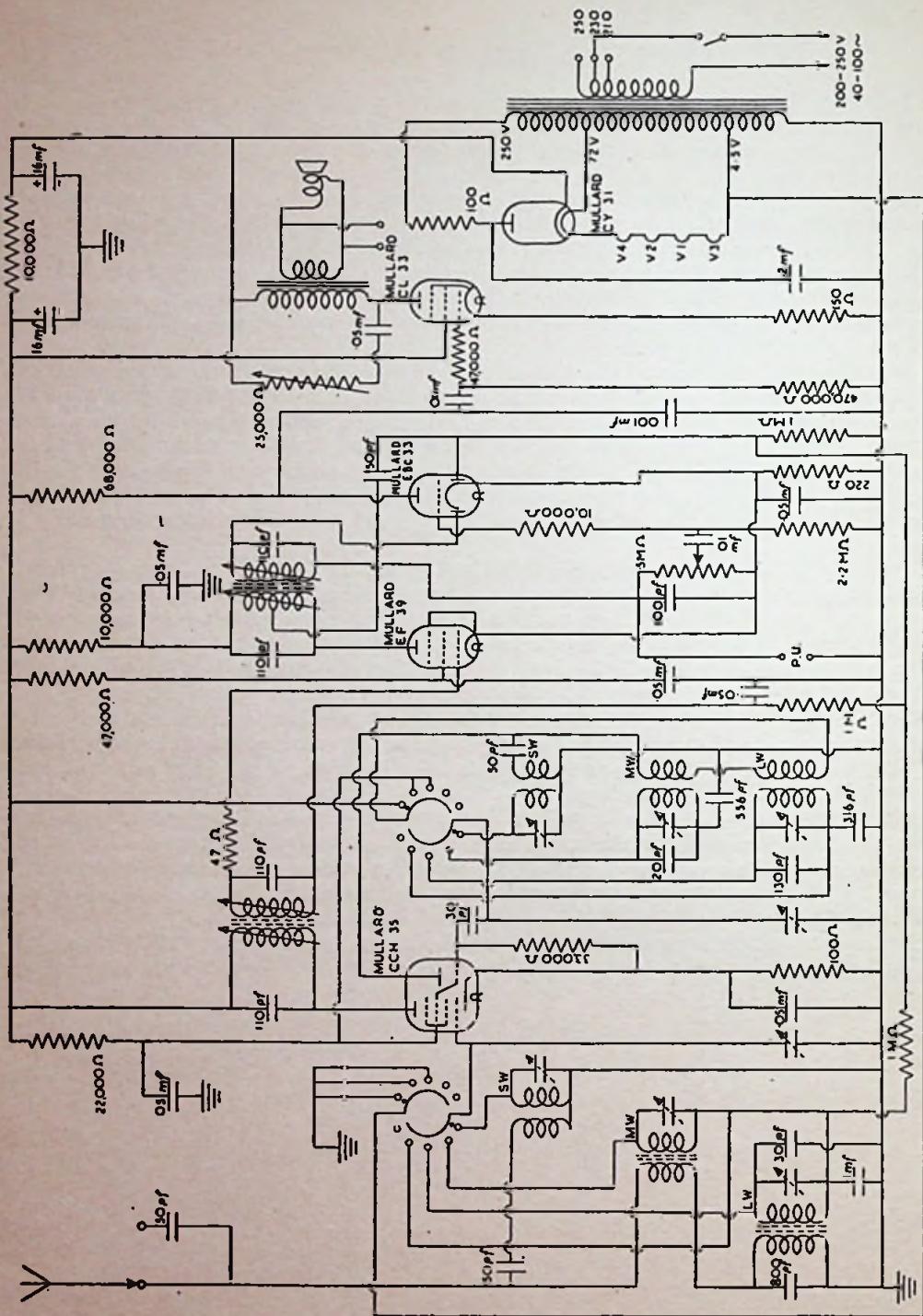
Commercial circuits which form the subject of pages 2-18 are followed by thirteen representative circuits ; four of these circuits are rearrangements of those published in *Modern Practical Radio and Television*, Chapter 5, Volume II, with component values added. The rearrangement has been undertaken, since the original circuits were designed for the sole purpose of forming a basis for comment, and many functions were deliberately duplicated for the purpose of broadening discussions. These rearranged circuits can be regarded as an individual interpretation of the original circuits, and form a most interesting subject for comparison.

In the various circuits illustrated in the following pages values have, of necessity, been specified, since in certain cases component values are peculiar to the valves selected. The author desires to make it clear, however, that he considers valves made by any member of the British Valve Association to be satisfactory providing, of course, they are associated with suitable component values. In short, the selection of valves in the following pages has been almost entirely influenced by the types in the author's possession.

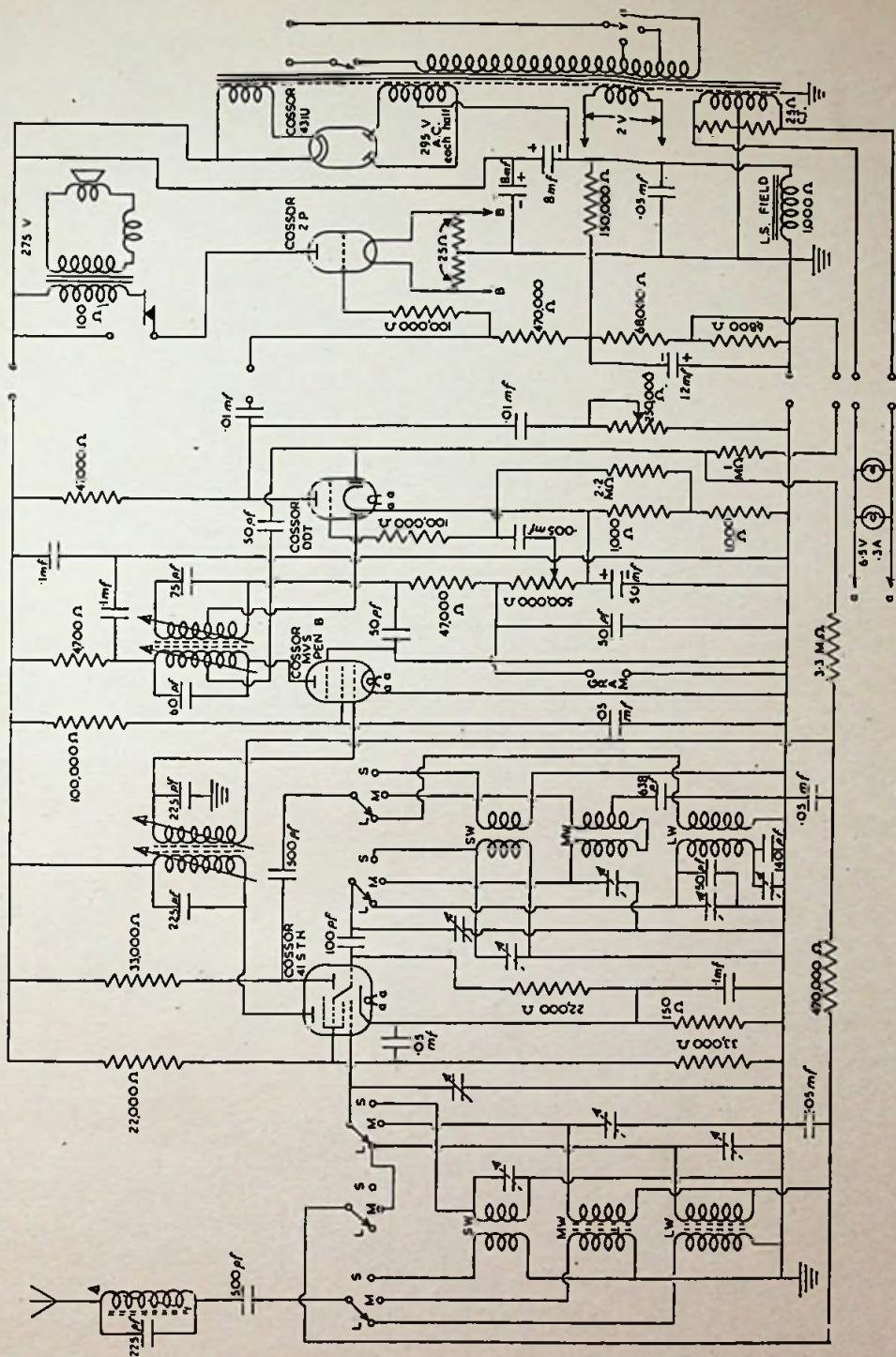
Pages 32-42 contain a variety of "unit" circuits useful when modifying or adding to existing equipment, and much thought and trouble has been expended in selecting circuits that will have wide appeal. One or two circuits have, however, been added on account of their intrinsic interest and originality.

Component values have been specified in these "unit" circuits wherever possible. Where values are omitted, it is because they are influenced by the complete circuit of which these "unit" circuits are intended to form only a small part ; in nearly every case, however, the missing value can be determined from studying the fundamental principles involved in the relevant section of *Modern Practical Radio and Television*.

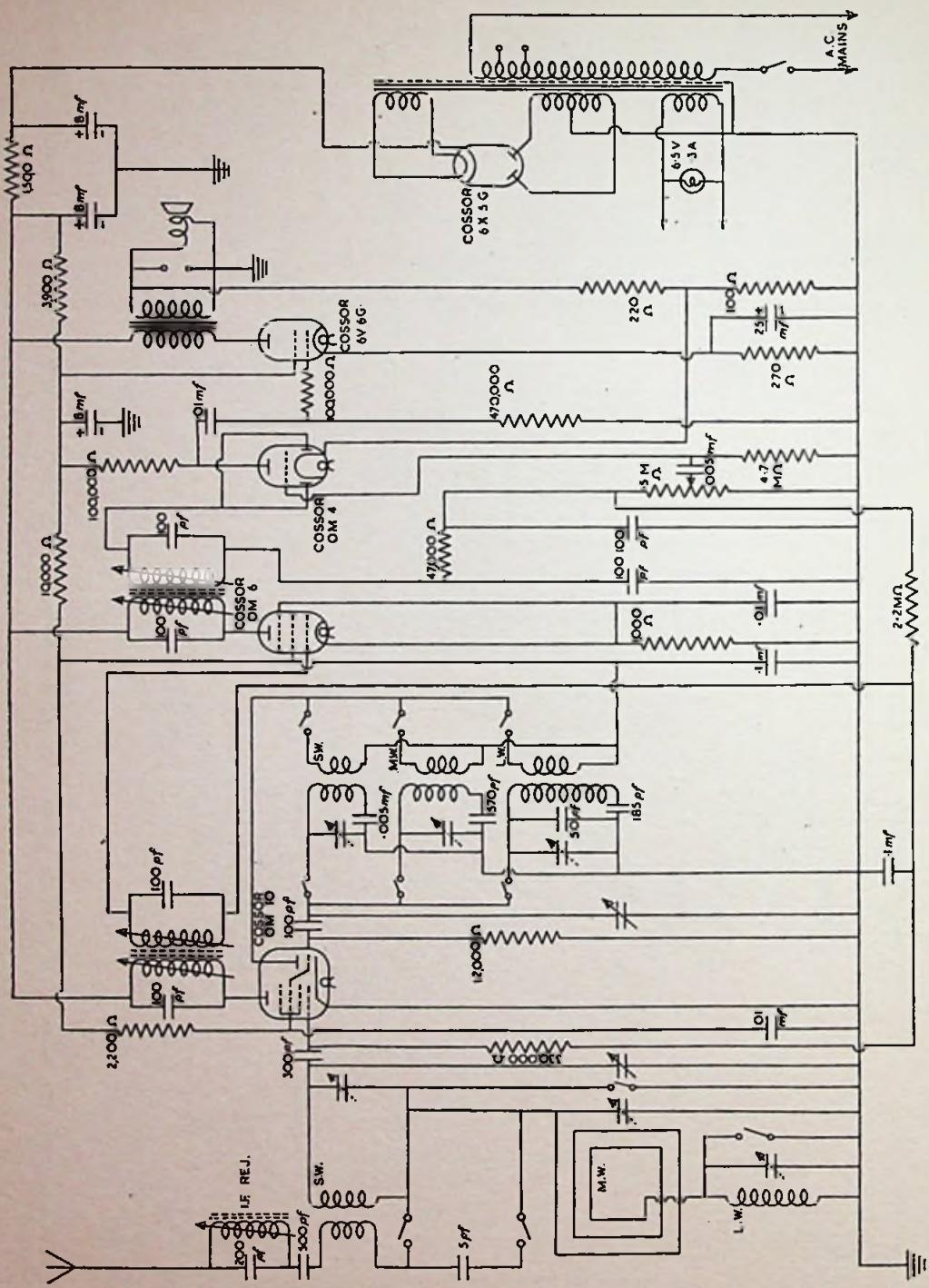
The author welcomes suggestions for "unit" circuits, or indeed any other items for possible inclusion in future editions of this data book.



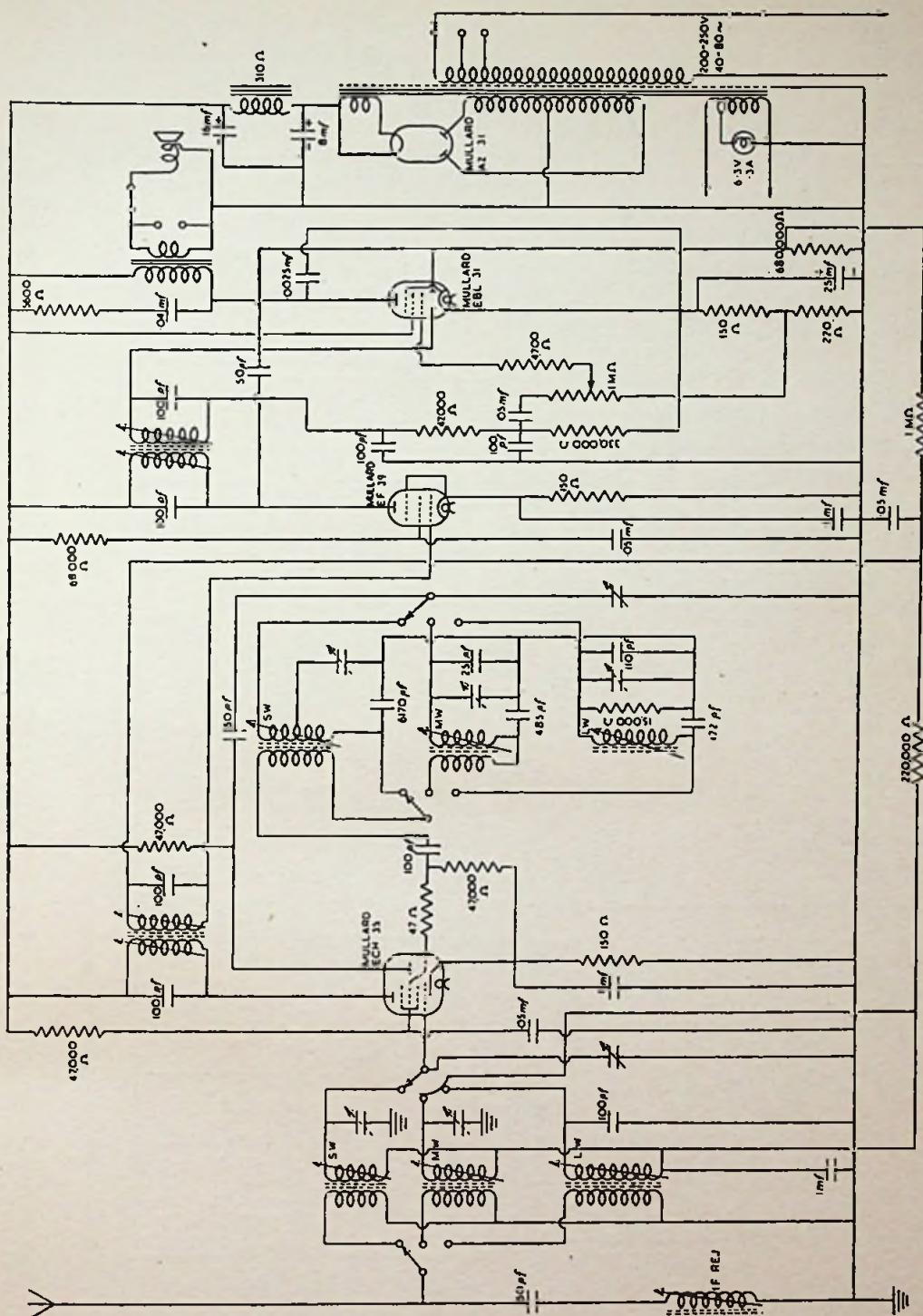
TR2 Bush Model A.C.81
 Attention is drawn to the output stage in which negative feedback is introduced by the omission of a cathode bypass condenser; note that the output valve anode is fed from the unsmoothed side of the H.T. supply, also the parasitic stopper in the I.F. amplifier grid lead.



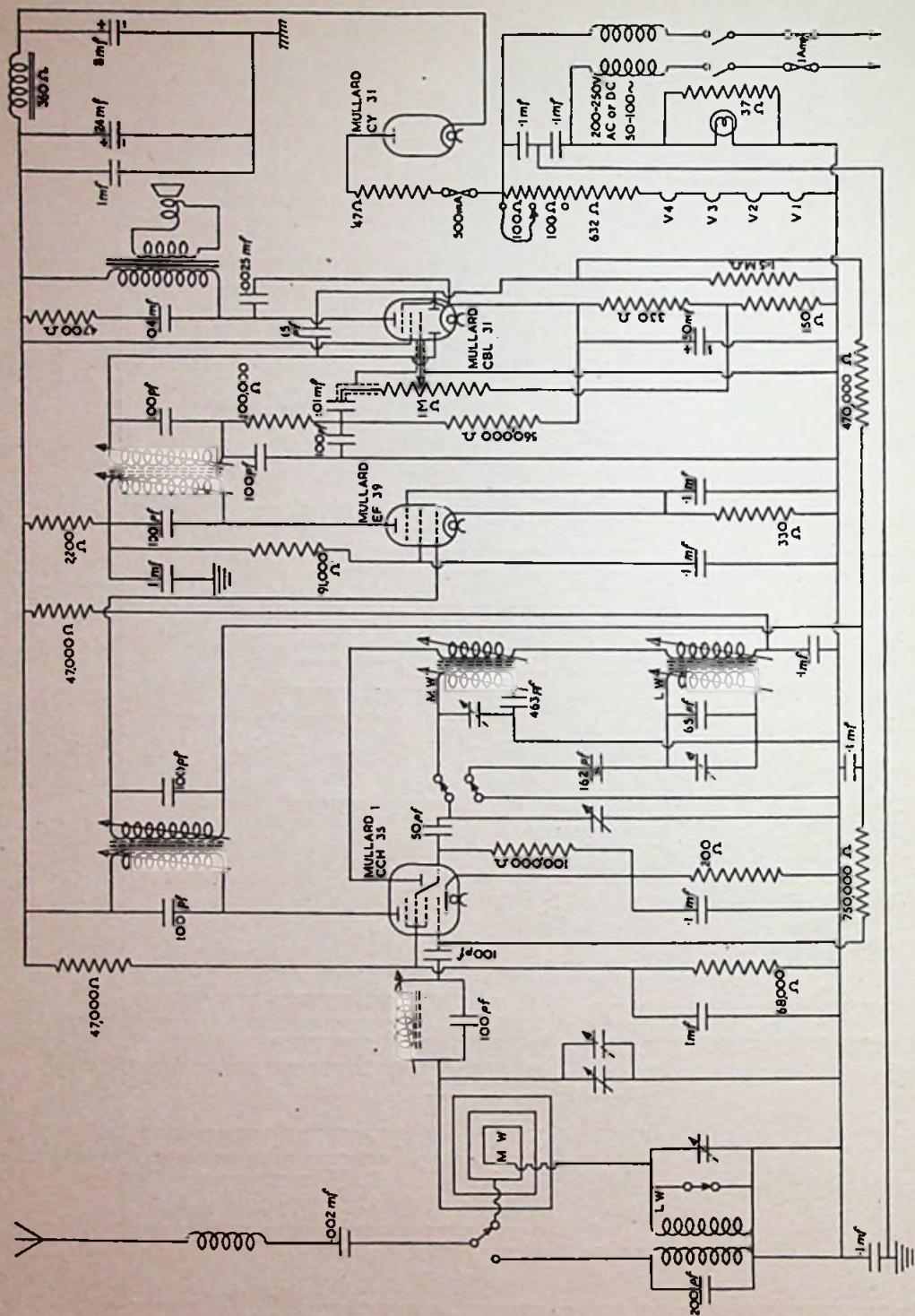
The Coscor Model 460 A.C.
Note the directly heated low filament voltage output valve, the bias for which is developed across the loudspeaker field coil placed in the negative H.T. lead.



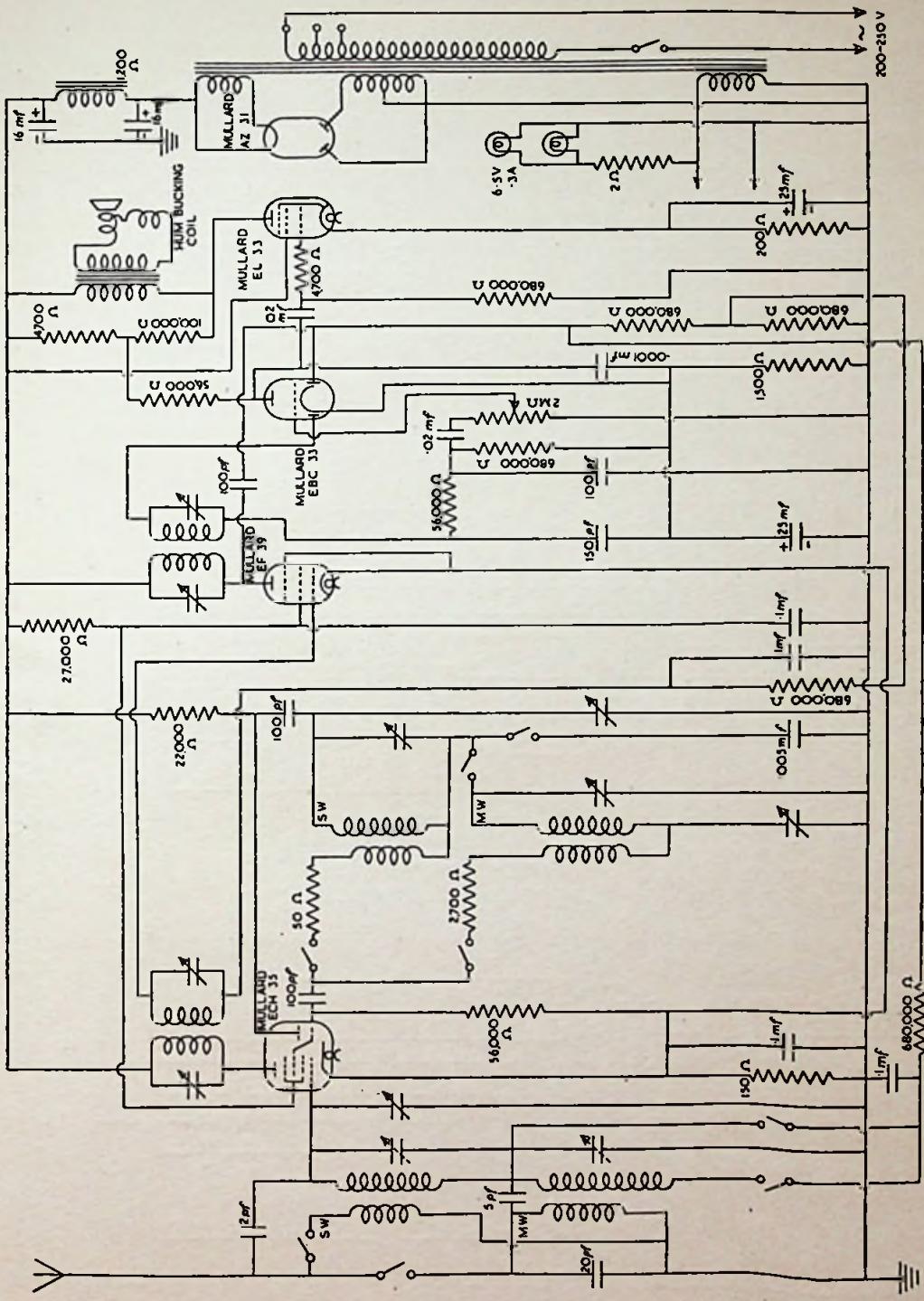
THE Cossor Model 404
The serial circuit arrangements are interesting and are worthy of careful study. Attention is also directed to the method of applying negative feedback and the resistance capacity two-stage H.T. smoothing.



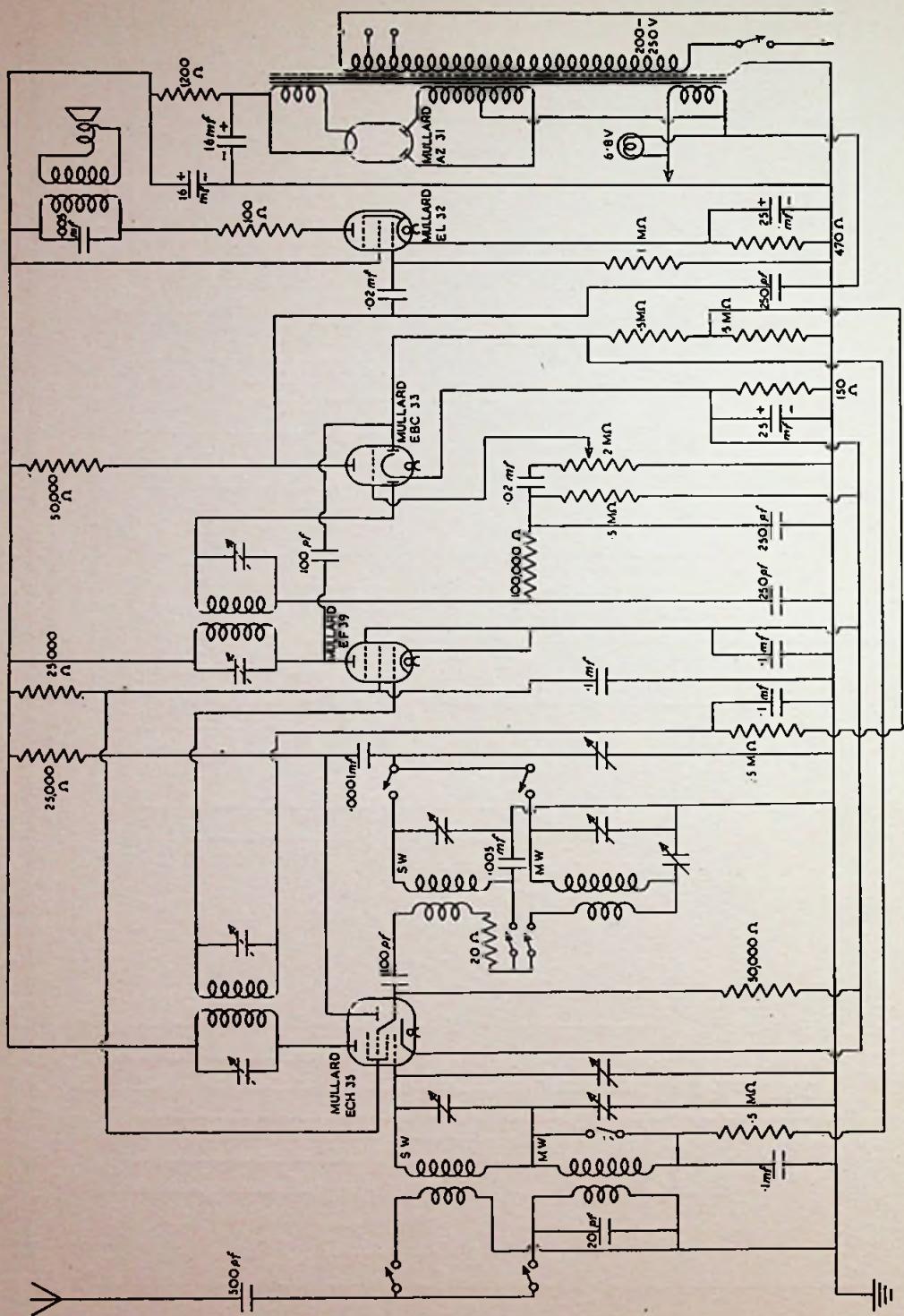
THE EXCO MODEL A22
Points of interest are the oscillator coupling and the method of obtaining negative feedback.



THE ERCO MODEL U29
 Note the rejector inductance in the aerial lead and the position of the I.F. rejector circuit between the frequency changer grid and aerial circuit. Observe that fuses are provided in both mains leads and rectifier anode lead, observe also the mains filter.

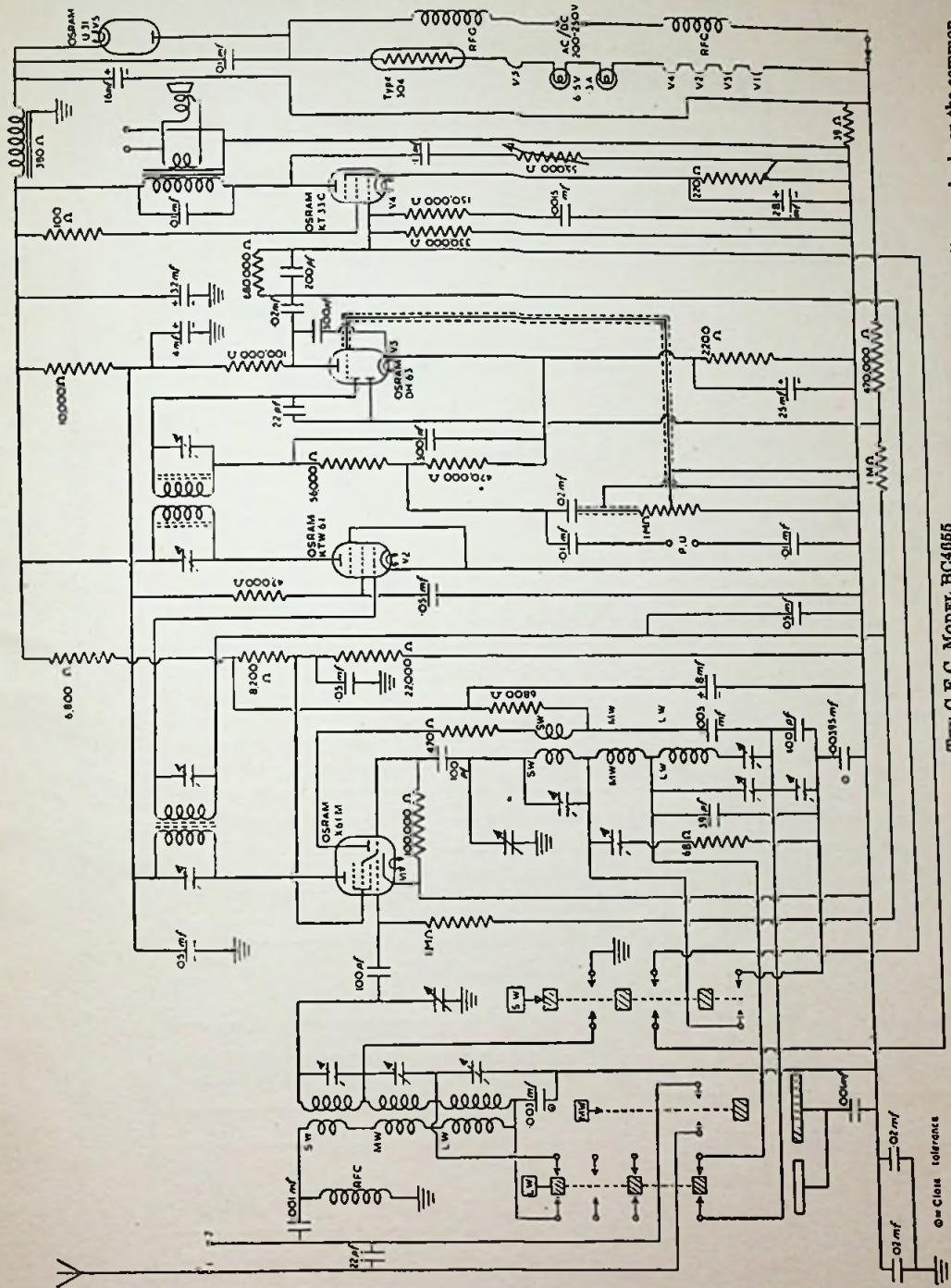


THE FERGUSON MODEL 450
Points of interest are the use of very small condensers in the design of the aerial couplings, anti-squeegging resistance in the oscillator grid circuits and the series negative feedback arrangements.

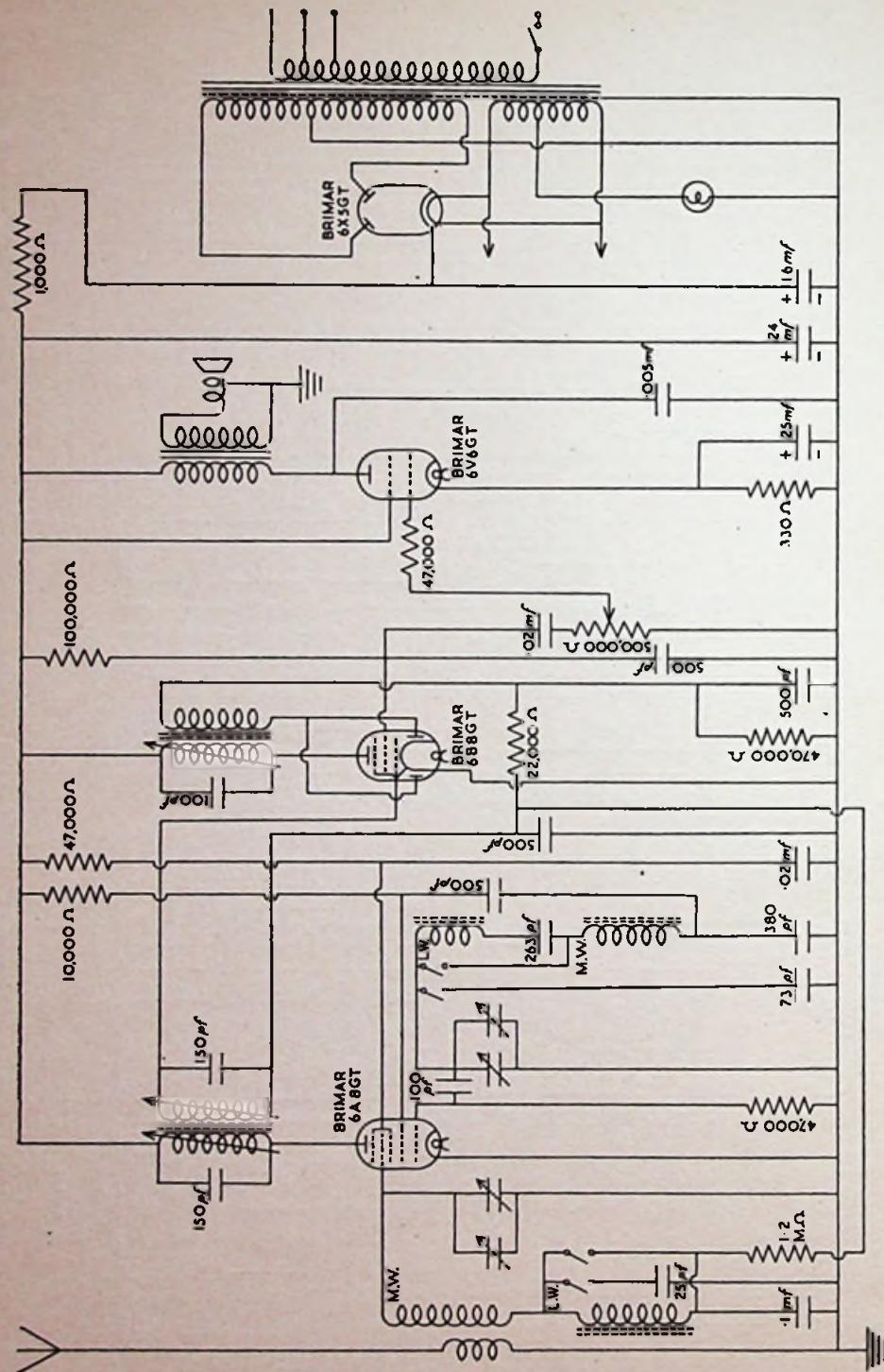


THE FERGUSON MODEL 910

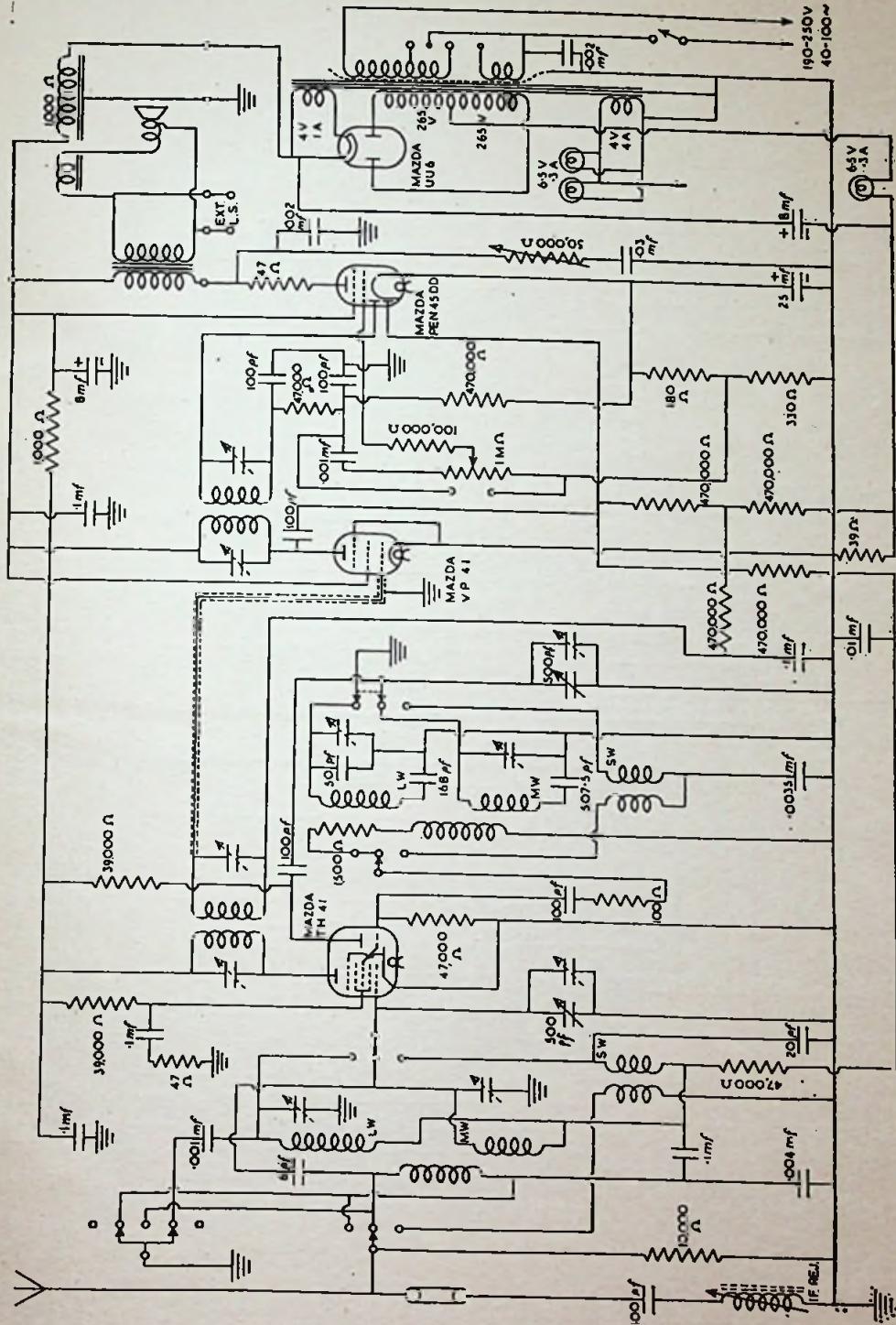
Points to note are parasitic stoppers in the oscillator grid and output anode circuits, resistance capacity smoothing and the fact that A.V.C. is derived from the anode of the I.F. amplifier to reduce sideband shrink when tuning.



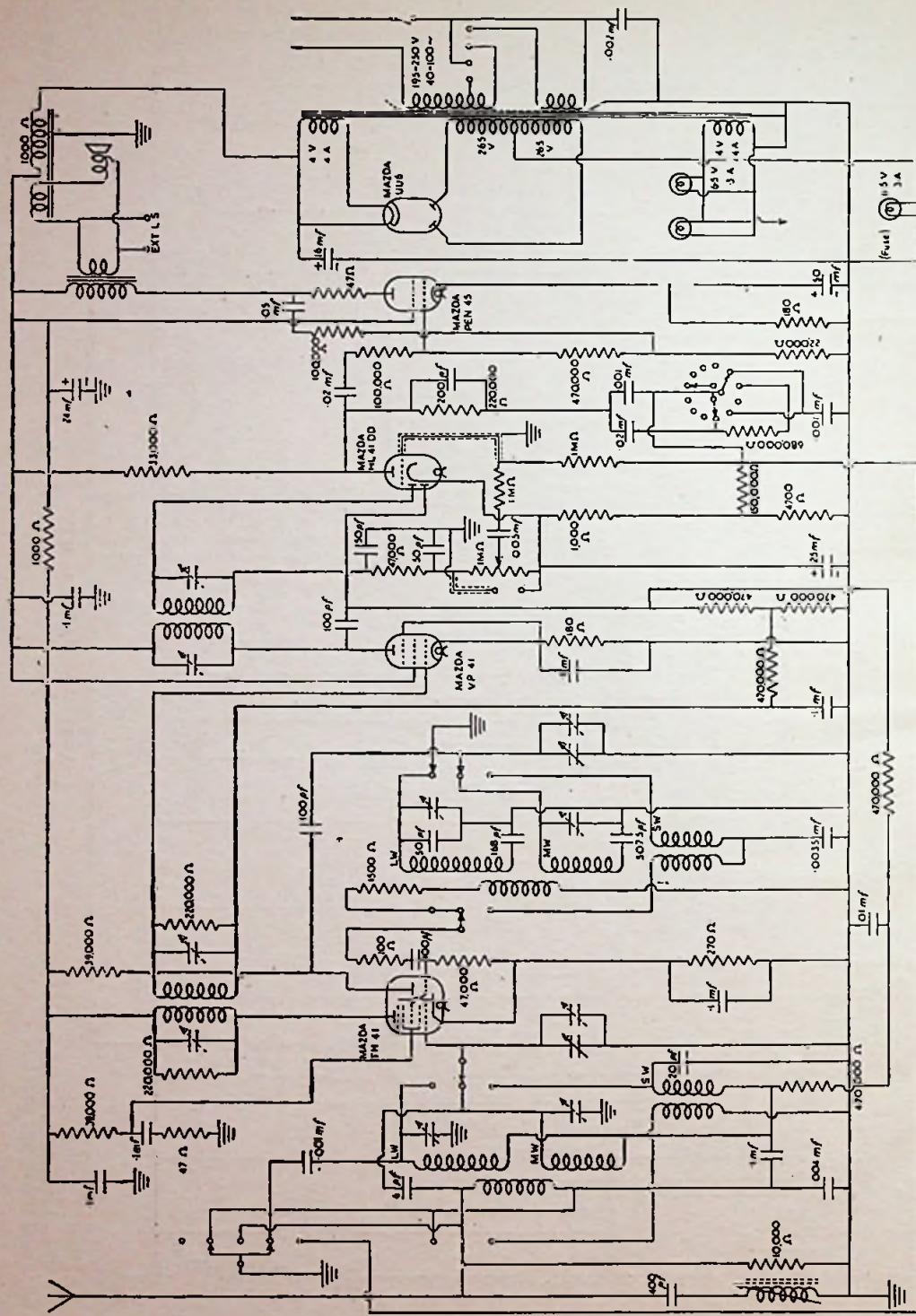
TUNING
 Tuning G.E.C. Model HC4065
 Both serial and oscillator coupling is unusual, the medium and long wave sections have only single turn primary windings and rely on the common impedance of the condensers marked close tolerance. Attention is also drawn to the R.F. choke in the aerial circuit which eliminates any tendency to modulation hum.



The KOLSTER-BRANDES BM20
Note particularly the middle stage, which is reflexed, being an I.F. amplifier and a L.F. amplifier; the screen acts as the anode for the latter function.

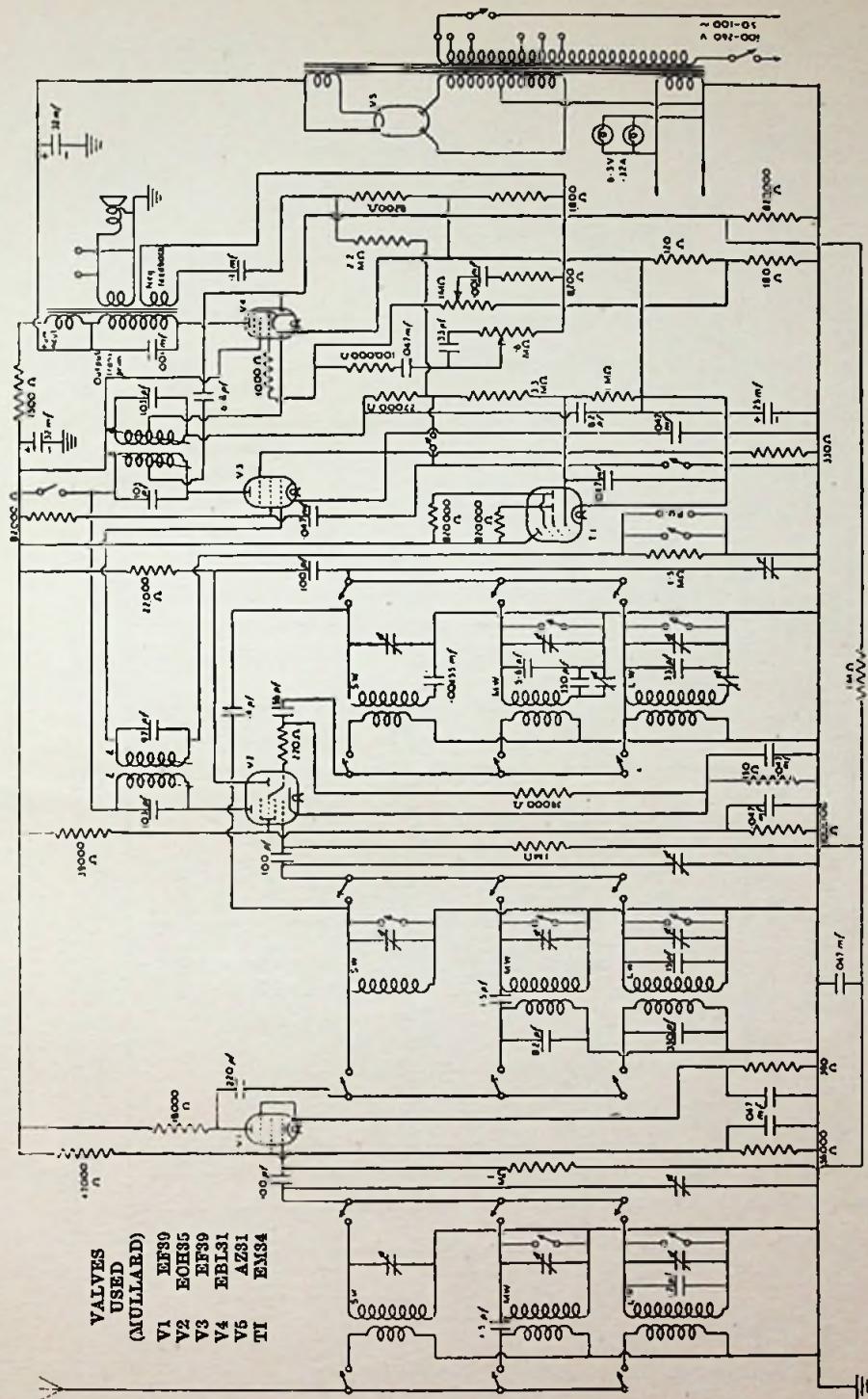


The McMichael Model 451 circuit, but attention is directed to the serial circuit and switching, the output valve grid circuit and the mains transformer primary.

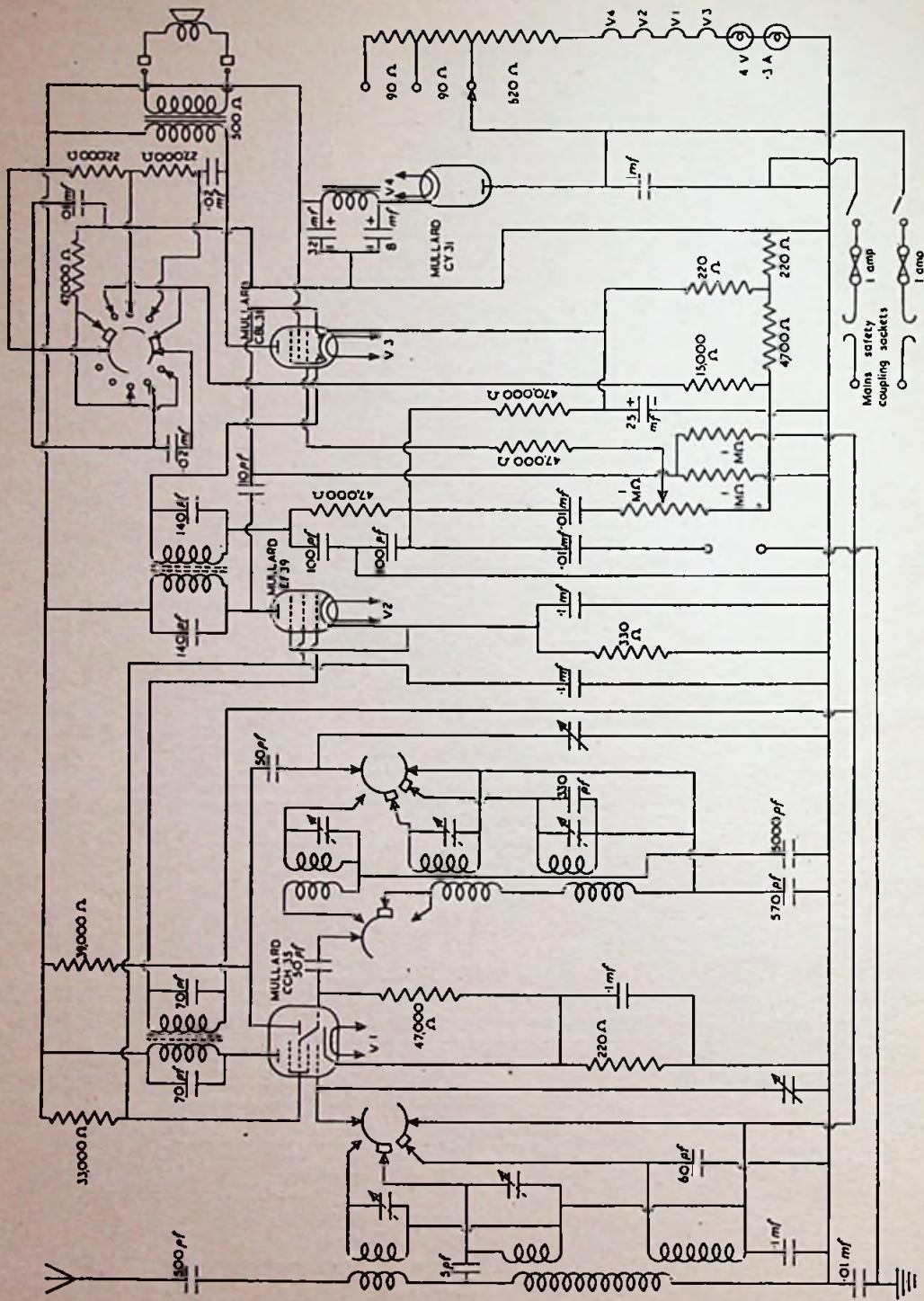


The McMichael Model 452

Note parasitic stoppers in the frequency changer screen lead and output valve anode lead, also anti-squeegging resistance in the oscillator grid circuit.
The resistance condenser combinations given by the four-position tone control are of interest; note that negative feedback is employed.

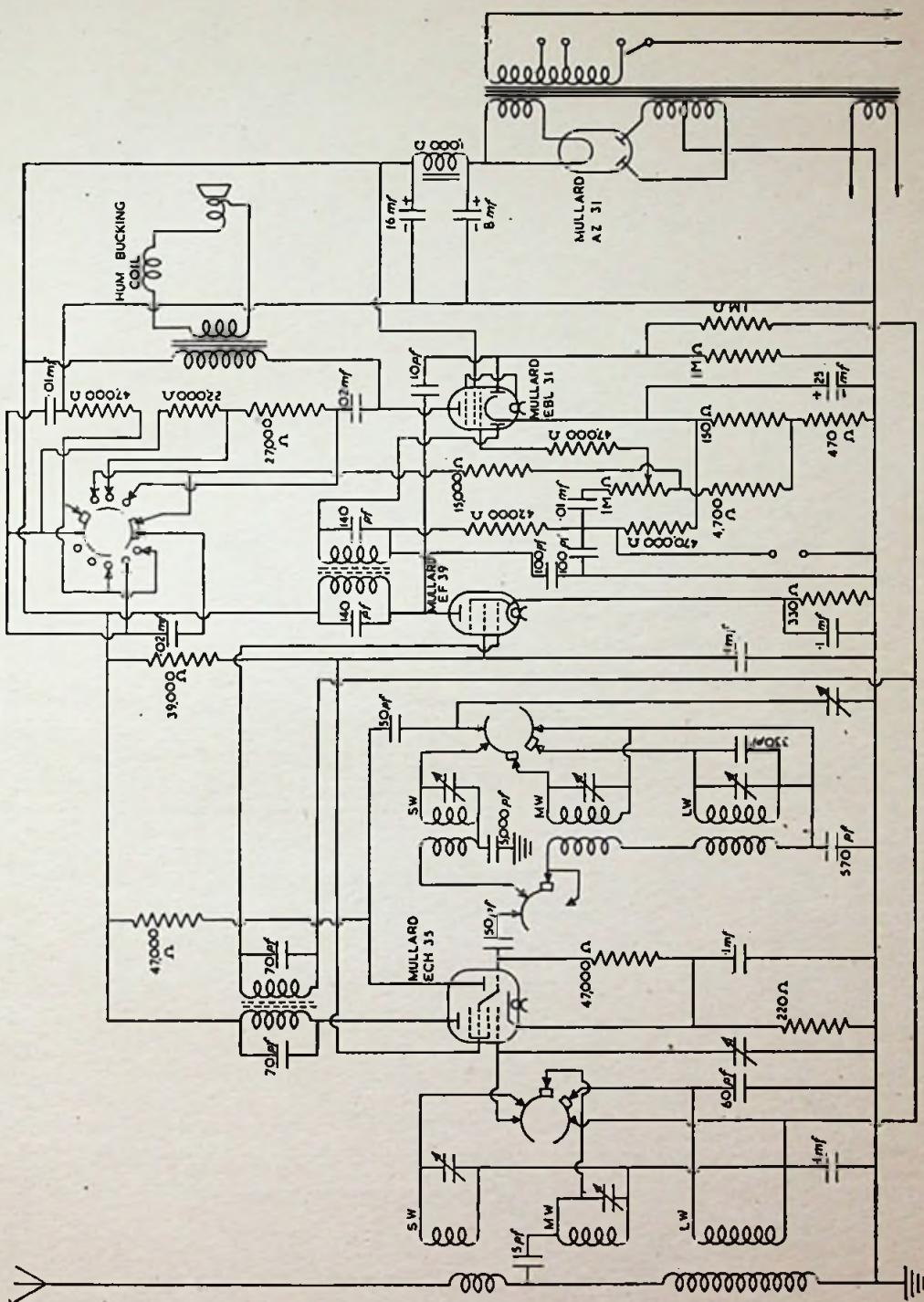


THE PHILIPS MODEL 170A
The negative feedback arrangements are an interesting object lesson in circuit design; the volume control is compensated by automatically varying negative feedback and the tone control operates by varying the amount of negative feedback at higher frequencies. Note also the hum neutralising arrangements.

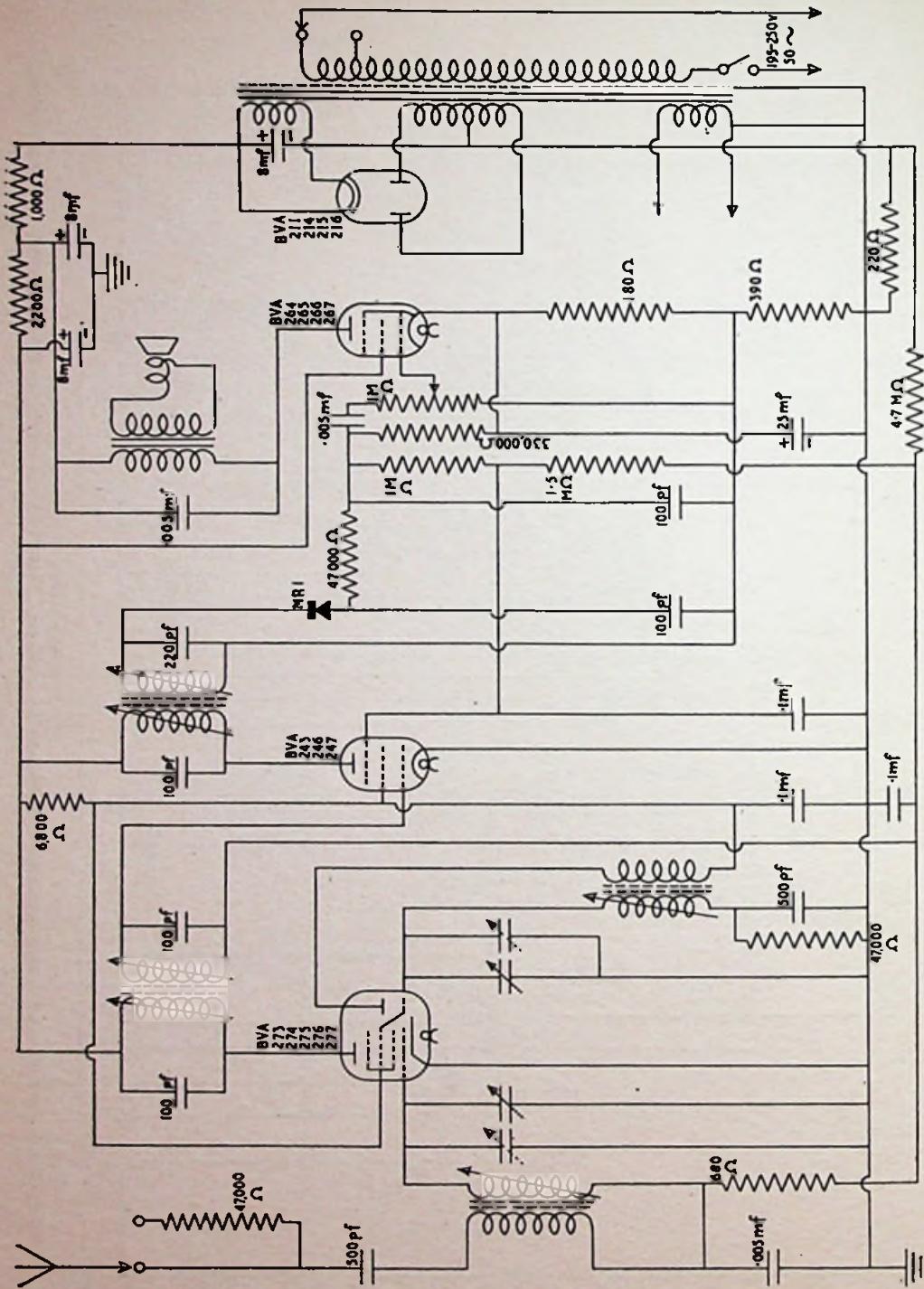


THE PYE MODEL 45A

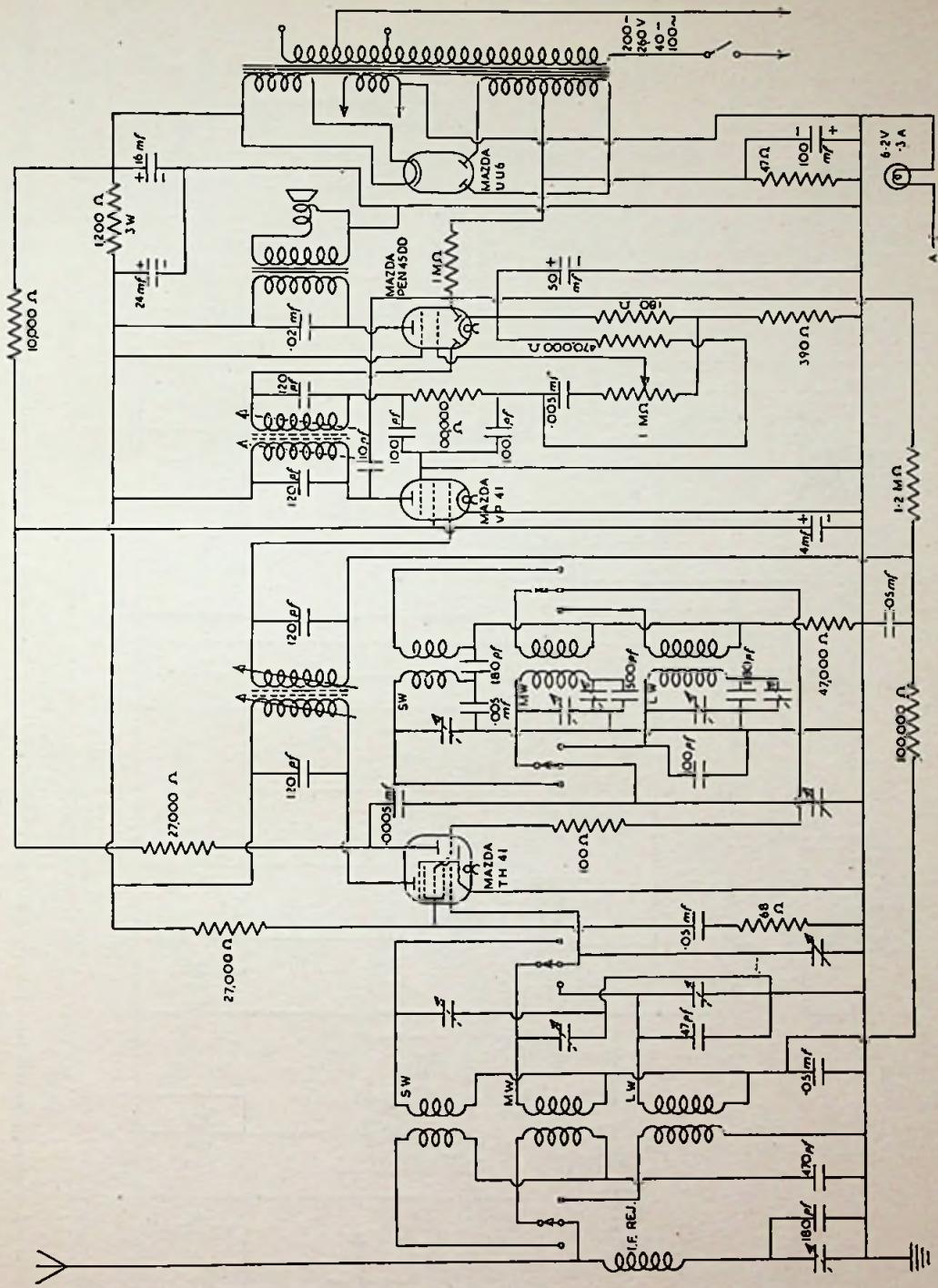
Like the model 15A, attention is drawn to the tone control arrangements. These two circuits form an interesting comparison between A.C. and D.C./A.C. working.



Attention is directed to the tone control switch, which has four positions. It is well worth following out the connections for each position.
THE PYE MODEL 16A

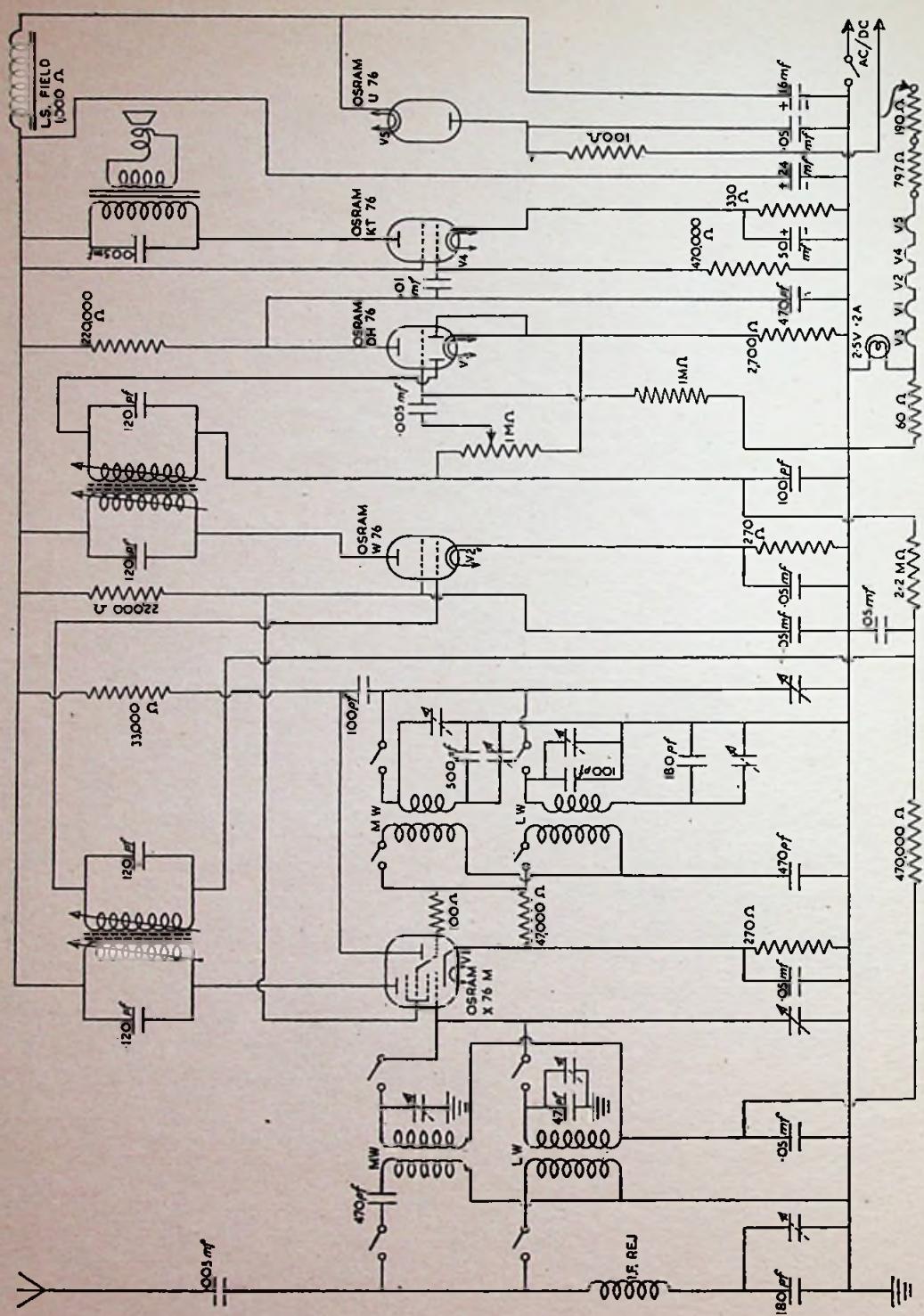


STANDARD WAR-TIME A.C. RECEIVER
This circuit is included as an example of austerity and because of certain ingenious features, notably the use of the I.F. amplifier suppressor (acting in conjunction with the cathode as a diode) to produce A.V.C. The MRI is a metal oxide detector. This receiver was manufactured by a number of firms who introduced sundry variations.



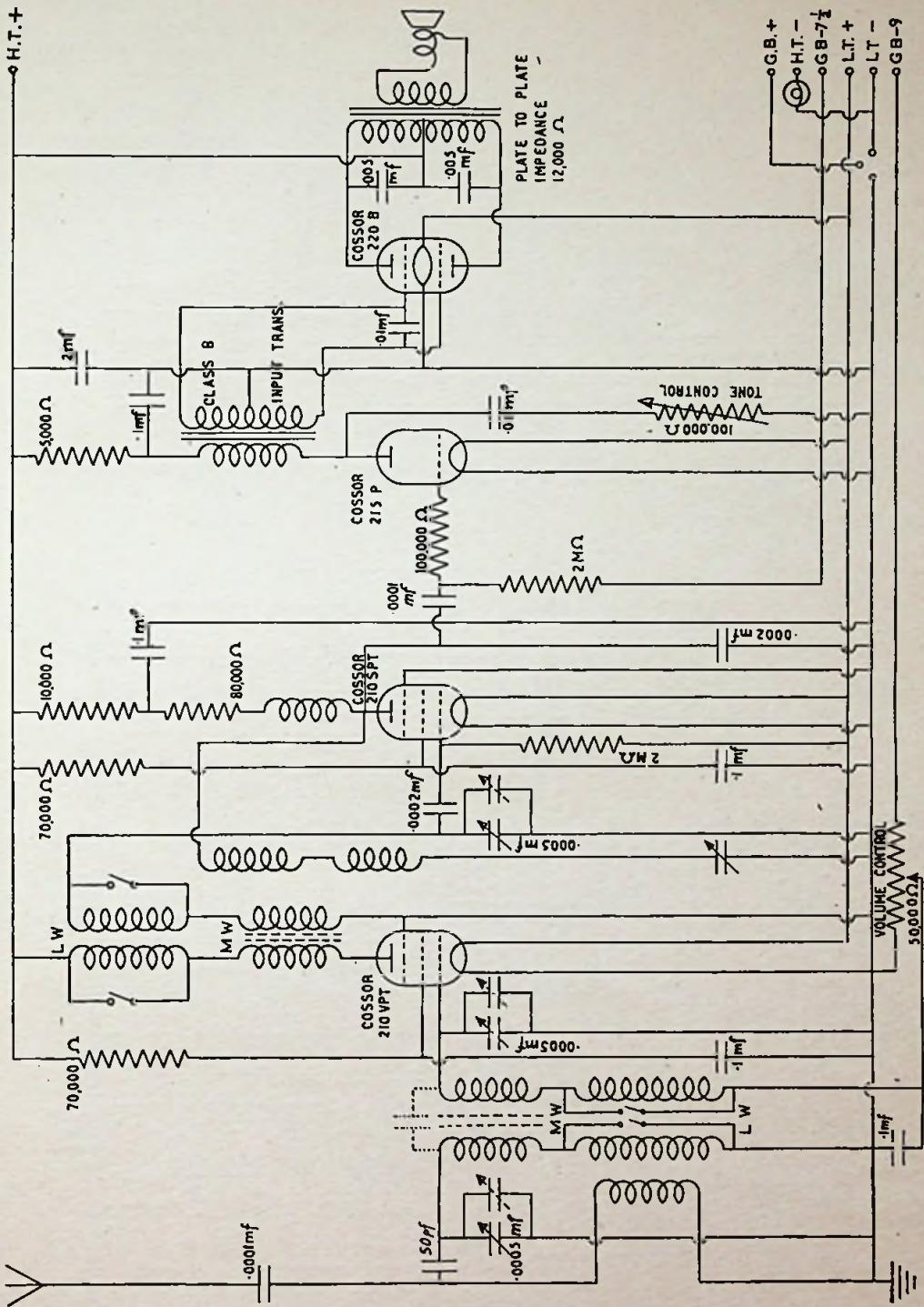
TUNA ULTRA MODEL T401

Note parasitic stoppers in the frequency changer screen lead and anti-equilibrating resistance in the oscillator grid lead. The biasing of the frequency changer and L.F. amplifier and the A.V.C. delay arrangements are additional points of interest.

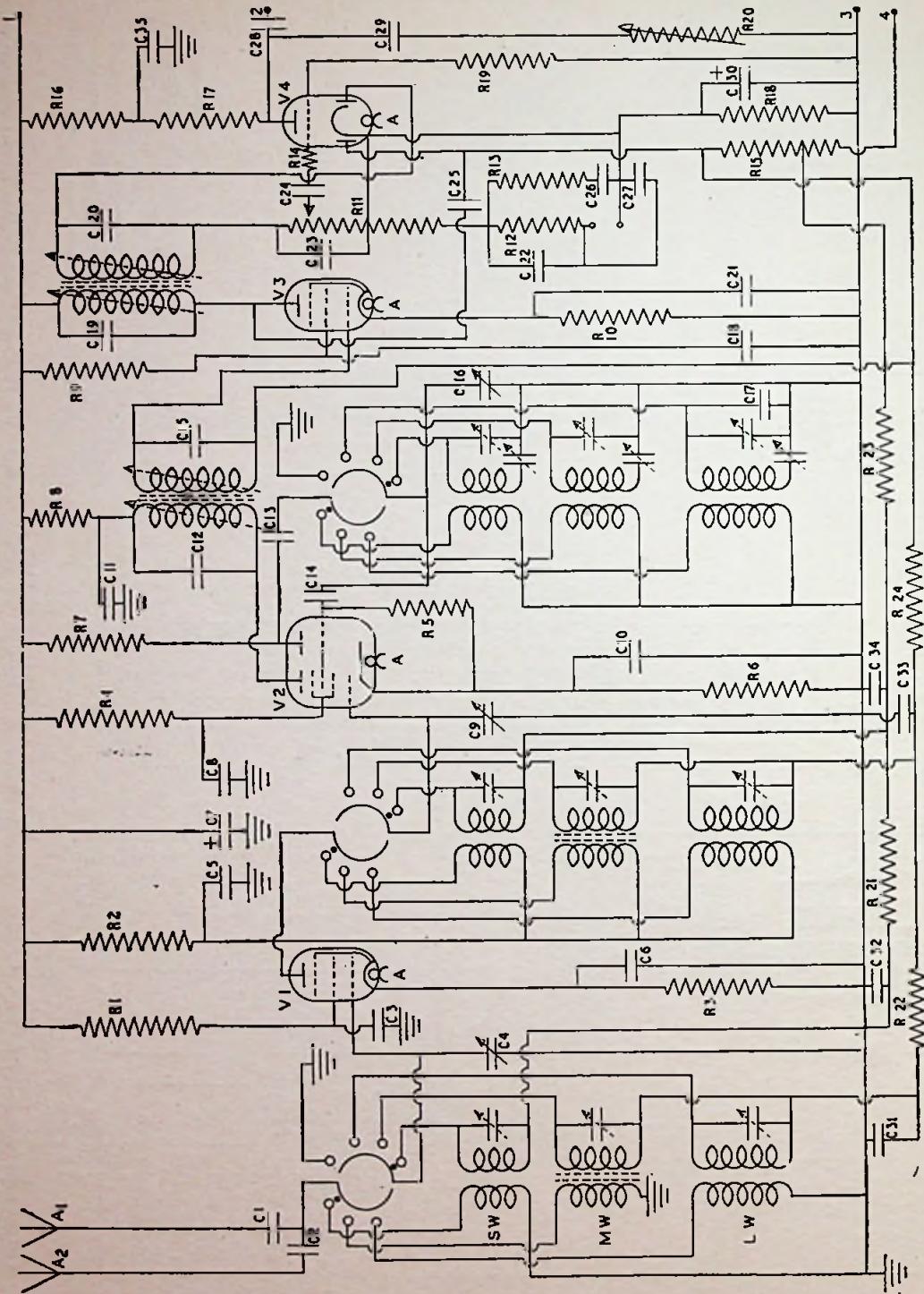


THE ULTRA MODEL U405

Note the anti-squegging resistance in the oscillator grid and the deliberate omission of the by-pass condenser across V3 cathode, introducing negative feedback.



A slight rearrangement of the class B receiver, shown in Chapter 5, Volume II, but with component values added. The aerial circuit uses bandpass coils which sometimes require small additional capacitive coupling, usually about 5 pf, shown dotted in the above illustration.

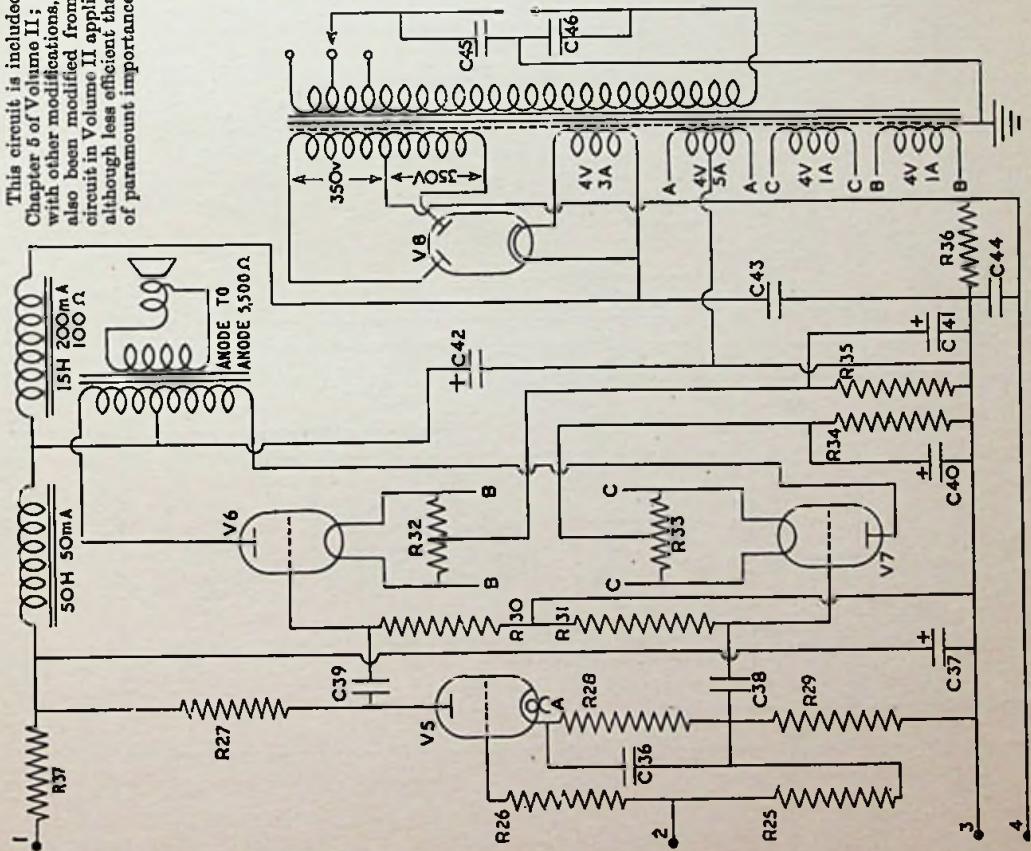


7 + 1 VALVE, 3-BAND RECEIVER
The output section and component values are on the facing page.

7 + 1 VALVE, 3-BAND RECEIVER

(12 watt, push-pull output)

This circuit is included in response to numerous requests for more details of the circuit shown in Chapter 5 of Volume II; it is presented, however, without the automatic frequency control circuit and with other modifications, so that non-standard components have been excluded. The output stage has also been modified from 8 to 12 watts. Much of the text matter which accompanies the 10-valve circuit in Volume II applies equally to the rearrangement herewith. Triode output valves are used, and although less efficient than tetrodes, they still enjoy considerable popularity where H.T. current is not of paramount importance.



CONDENSERS		CONDENSERS	
C1 = 100 pf	C29 = -0.2 mf	C29 = -0.2 mf	R10 = 15,000
C2 = 15 pf	C30 = .50 mf	C30 = .50 mf	R10 = 100
C3 = -1 mf	C31 = -.05 mf	C31 = -.05 mf	R11 = -.5 + .5 MQ
C4 = TC*	C32 = -.03 mf	C32 = -.05 mf	R12 = 10,000
C5 = -1 mf	C33 = -.03 mf	C33 = -.05 mf	R13 = 3,300
C6 = -2 mf	C34 = -.03 mf	C34 = -.05 mf	R14 = 47,000
C7 = -2 mf	C35 = -.5 mf	C35 = -.5 mf	R15 = 1 MQ ¶
C8 = -2 mf	C36 = -1.5 mf	C36 = -1.5 mf	R16 = 20,000
C9 = TC	C37 = 16 mf	C37 = 16 mf	R17 = 56,000
C10 = -2 mf	C38 = -0.75 mf	C38 = -0.75 mf	R18 = 2,000
C11 = -2 mf	C39 = -0.75 mf	C39 = -0.75 mf	R19 = 1 MQ
C12 = IFTC †	C40 = .25 mf	C40 = .25 mf	R20 = 20,000
C13 = 100 pf	C41 = .25 mf	C41 = .25 mf	R21 = 1 MQ
C14 = 50 pf	C42 = 16 mf	C42 = 16 mf	R22 = 1 MQ
C15 = IFTC	C43 = 4 mf	C43 = 4 mf	R23 = 1 MQ
C16 = -1 mf	C44 = 1 mf	C44 = 1 mf	R24 = 1 MQ
C17 = Padder ‡	C45 = -1 mf	C45 = -1 mf	R25 = 130,000
C18 = -1 mf	C46 = -1 mf	C46 = -1 mf	R26 = 2,000
C19 = IFTC			R27 = 3,600
C20 = IFTC			R28 = 330
C21 = -1 mf	R1 = 15,000	R1 = 15,000	R29 = 3,600
C22 = -.01 mf	R2 = 6,100	R2 = 6,100	R30 = 100,000
C23 = 66 pf	R3 = 100	R3 = 100	R31 = 100,000
C24 = -.02 mf	R4 = 8,200	R4 = 8,200	R32 = 25
C25 = 75 pf	R5 = 33,000	R5 = 33,000	R33 = 25
C26 = -1 mf	R6 = 300	R6 = 300	R34 = 510
C27 = -.01 mf	R7 = 33,000	R7 = 33,000	R35 = 510
C28 = -.03 mf	R8 = 5,100	R8 = 5,100	R36 = 10
			R37 = 7,500

* Capacitance of variable tanking condenser will depend on limits of bands to be covered and inductance of coils selected.
† Capacity as required to tune selected I.F. transformers to intermediate frequency to be used.

‡ Capacity as required to raise total padding capacity to required figure.
§ Q36 may, with advantage, be connected across a portion only of R38 to provide a small amount of negative feed-back.

|| Values shown are preferred values: close non-preferred values can be used, with discretion.

¶ Tapped resistance gives any lesser degree of A.V.C. on shortwave band as desired. If very weak stations are required, move tap to top end of resistance and disconnect from grid of V1, when R3 would be 900Ω.

V1 Cossor MVS/Pen or MVS/Pan B

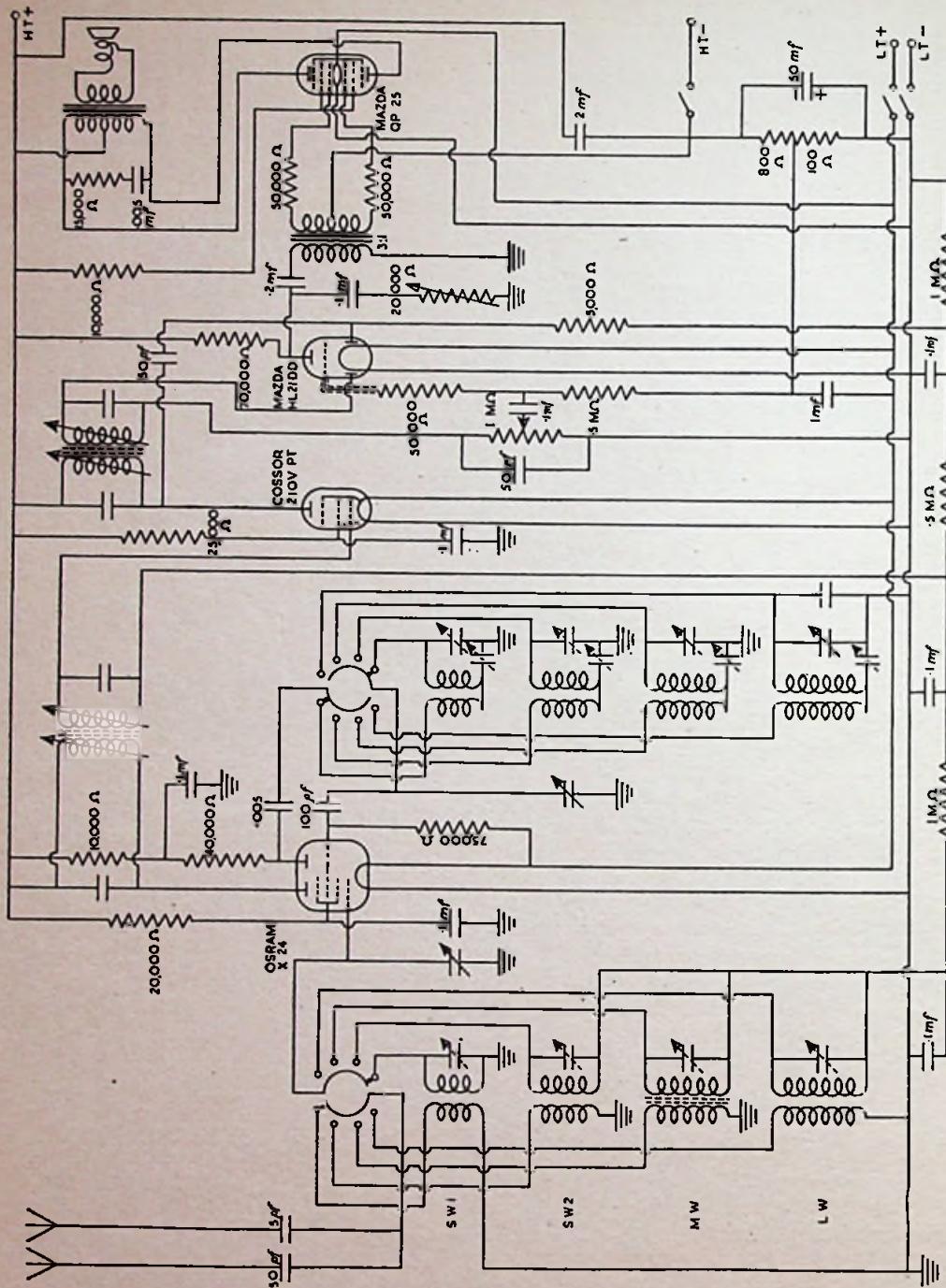
V6 Cossor 41MP **

V7 Mazda PP6/400

V3 Cossor MVS/Pen or MVS/Pan B

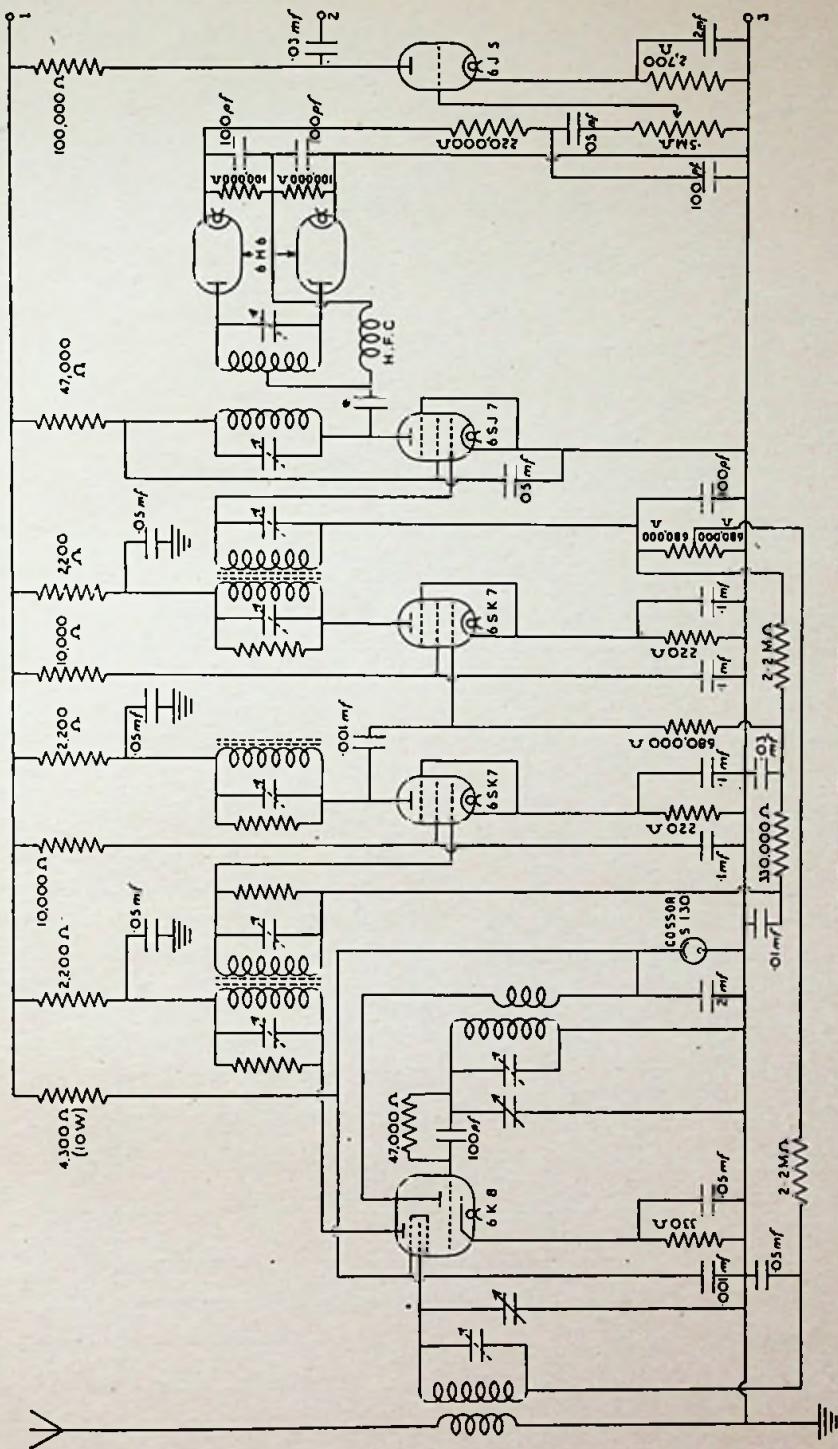
V4 Cossor DDT

** The author has used a valve of this type for several years with complete success under conditions prevailing in this circuit; nevertheless, a considerable voltage difference exists between heater and cathode. It is advantageous, therefore, to supply the heater from a separate transformer winding, the centre tap being connected to chassis through a 1 mil-condenser.



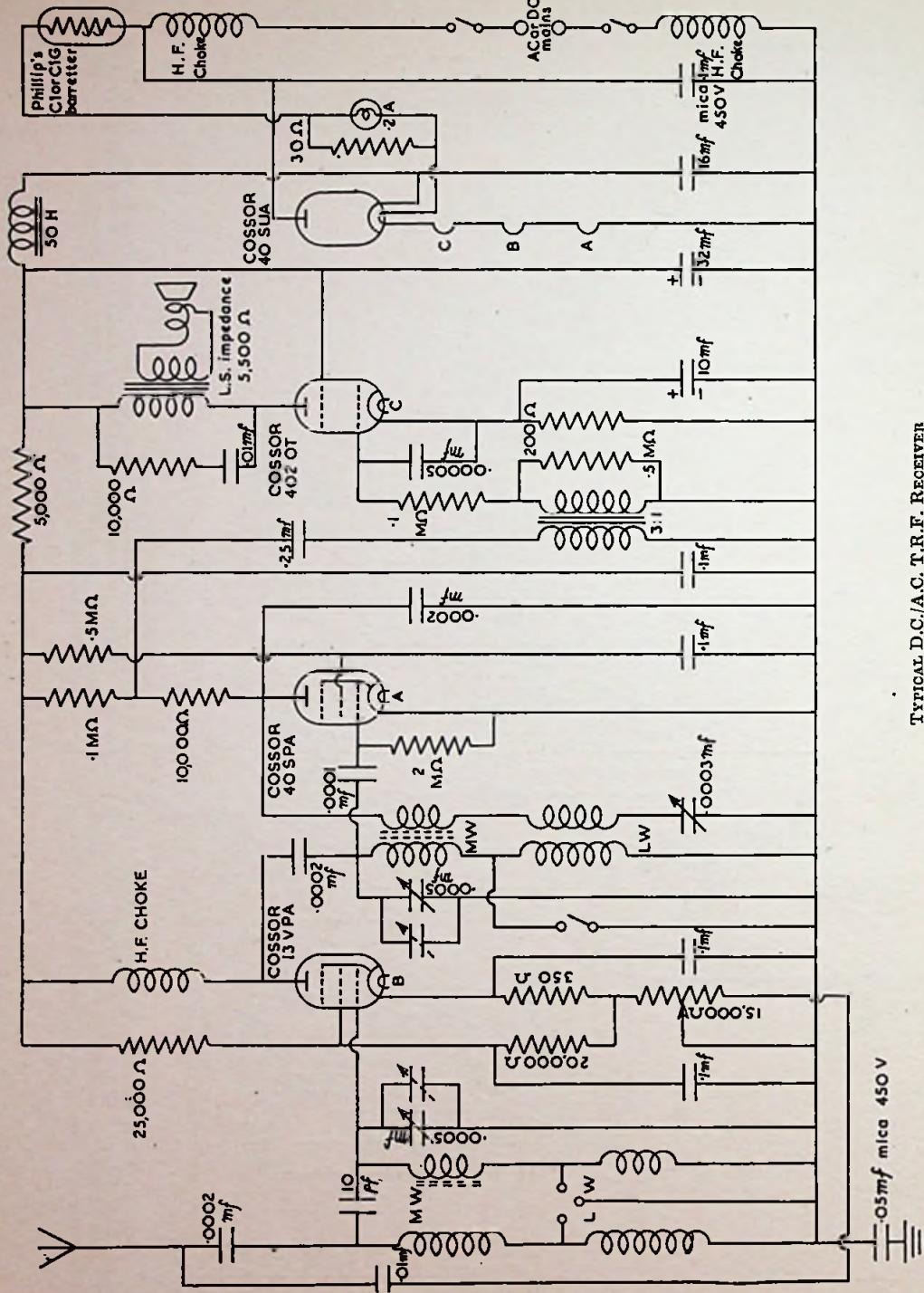
Q.P.P. SUPERHET BATTERY RECEIVER

A modification of the Q.P.P. receiver illustrated in Chapter 5, Volume II, but with typical values shown. Coils covering any desired number of wavebands can, of course, be used, and their selection will determine the values of the pre-set trimming and padding condensers. Should a frequency changer or I.F. amplifying valve be used of a type which passes grid current at zero grid volts, the lower end of the A.V.C. diode load resistance should be disconnected from L.T. — and connected to a point on the main bias resistance, which is just sufficiently negative to prevent the flow of grid current.



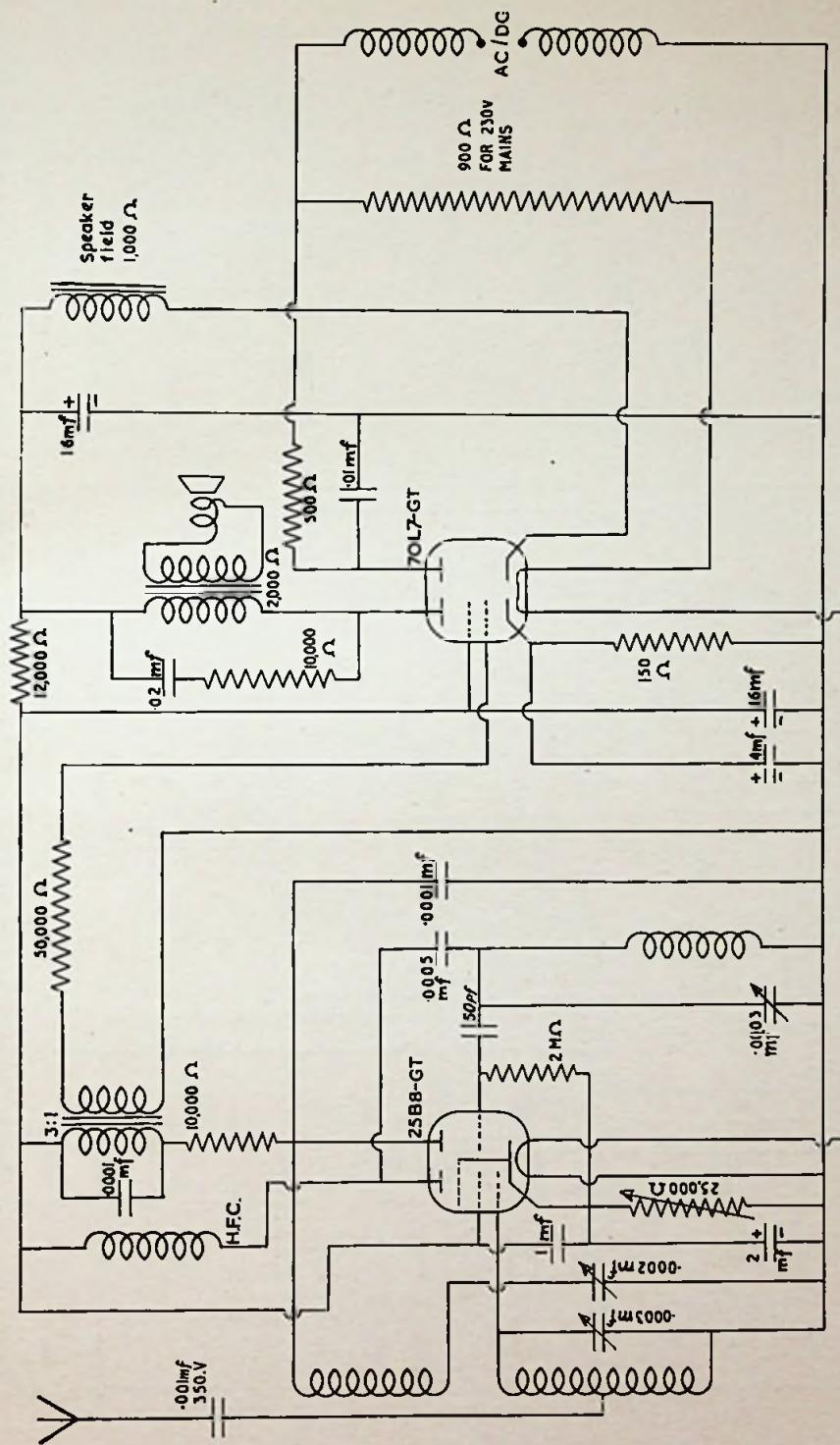
A TYPICAL WIDEBAND F.M. RECEIVER

This typical F.M. receiver is intended to be used with the output and power pack unit shown on page 21, which will need to be modified in the following way: Terminal 4, C₄₄ and R₃₆ are deleted and the points formerly bridged by R₃₆ joined together; the value of R₃₇ becomes 1,000 ohms. The H.F. and oscillator coils and the I.F. and discriminator coils will normally be part of a complete set designed for the waveband or bands to be covered; no value is shown for the condenser marked * as it will normally be an integral part of the discriminator coil assembly.



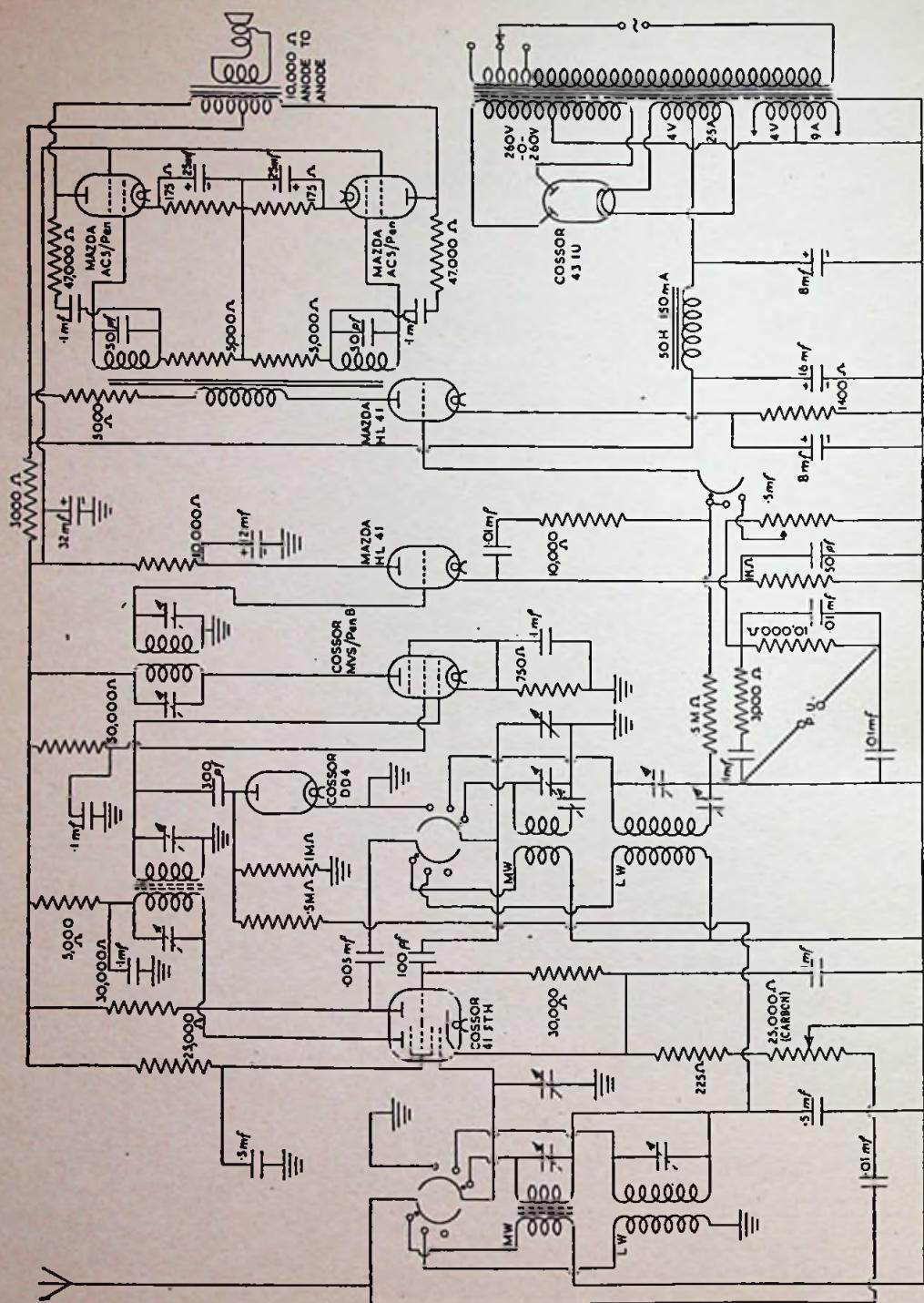
TYPICAL D.C./A.C. T.R.F. RECEIVER

This circuit is generally similar to the D.C./A.C. circuit in Chapter 5, Volume II, but values are shown and some modifications have been introduced. Variable negative feed-back could be introduced by using a potentiometer for the resistance across the L.F. transformer secondary, the slider being connected through a 1 mf. condenser to the output anode.

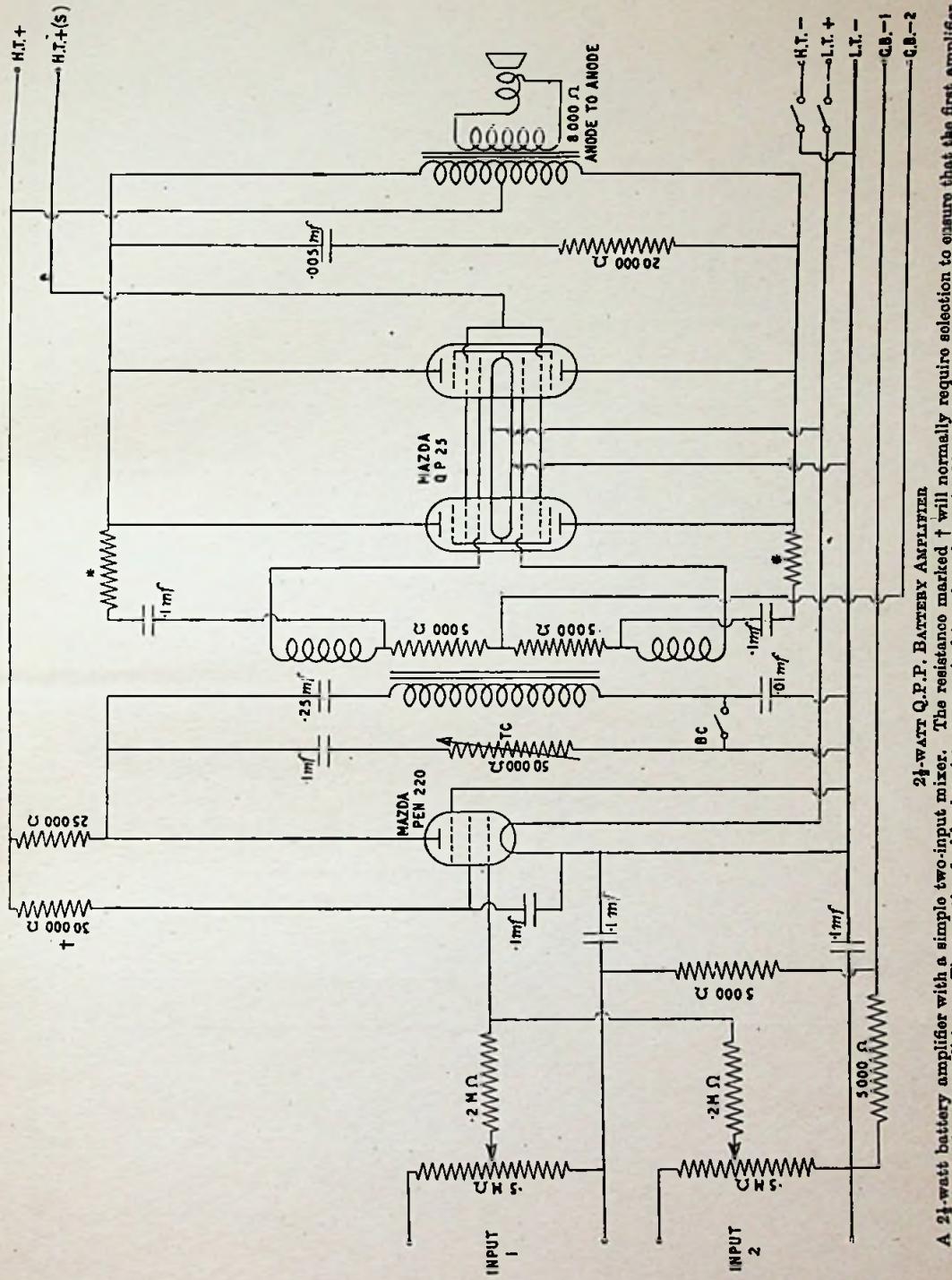


TYPICAL T.R.F. MIDGET RECEIVER

This circuit is typical of many of the cheaper midget receivers manufactured in the U.S.A., and although fundamentally a 3 + 1 valve receiver, only two valves are used, which are only 1½ inches diameter and 3½ inches long, including pins, making possible with the use of midget components a receiver of exceeding small dimensions. The rectifier anode dropping resistance (600 ohms) assumes a L.S. transformer primary resistance of 360 ohms; these considerations are important as the 70L7-GT has a maximum anode voltage of 117 volts and the 25B8-GT has a maximum anode voltage of 100 volts.

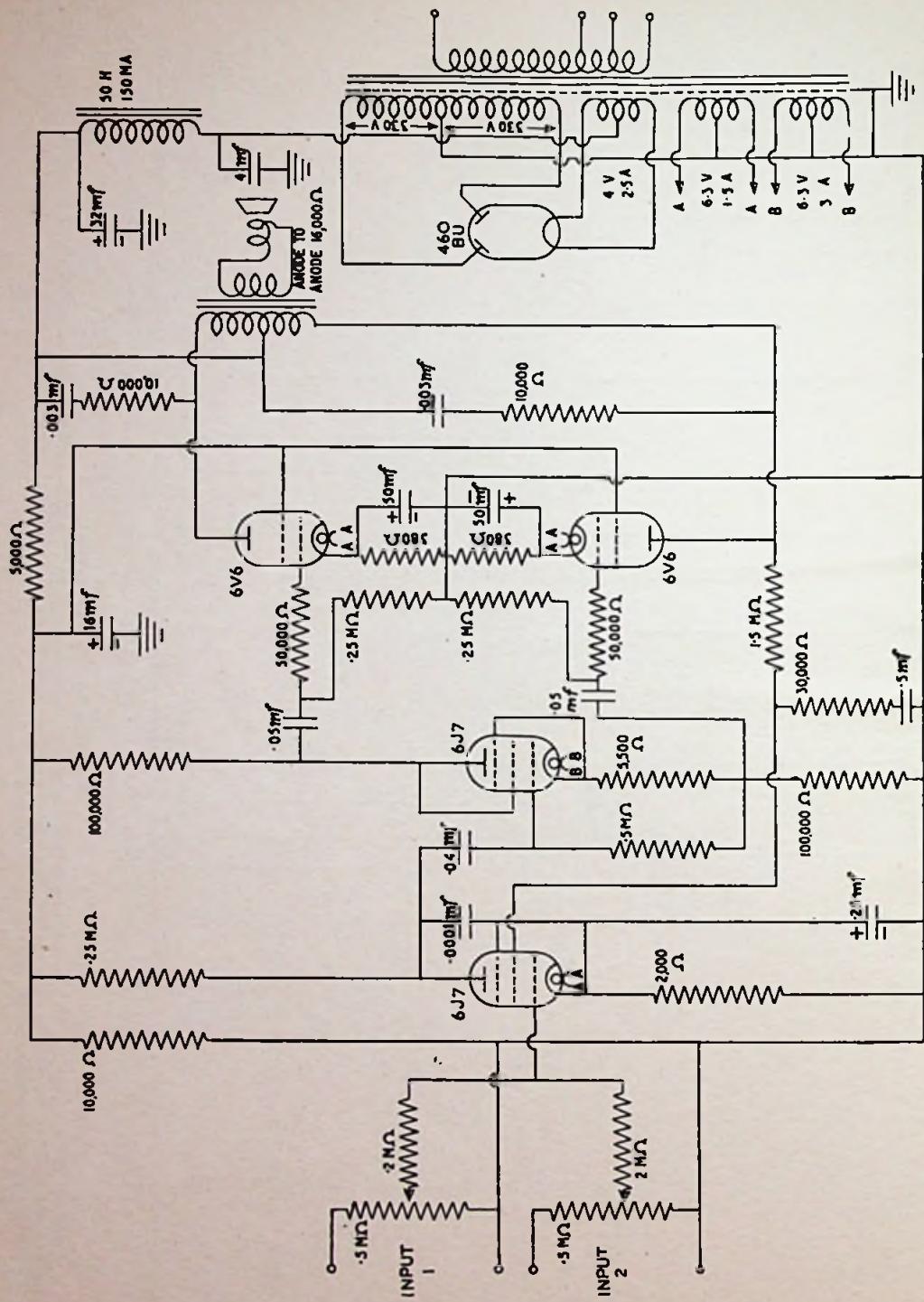


A MOST UNUSUAL RECEIVER
 Circuits presented in a book of this character are necessarily orthodox, and this circuit is introduced as a measure of relief and to illustrate that good results can be obtained in more ways than one. The circuit was found by the author in use in a recreation room, where it gave really excellent quality on more or less local stations. The circuit arrangement is probably the result of a series of modifications carried out without regard to the circuits as a whole.



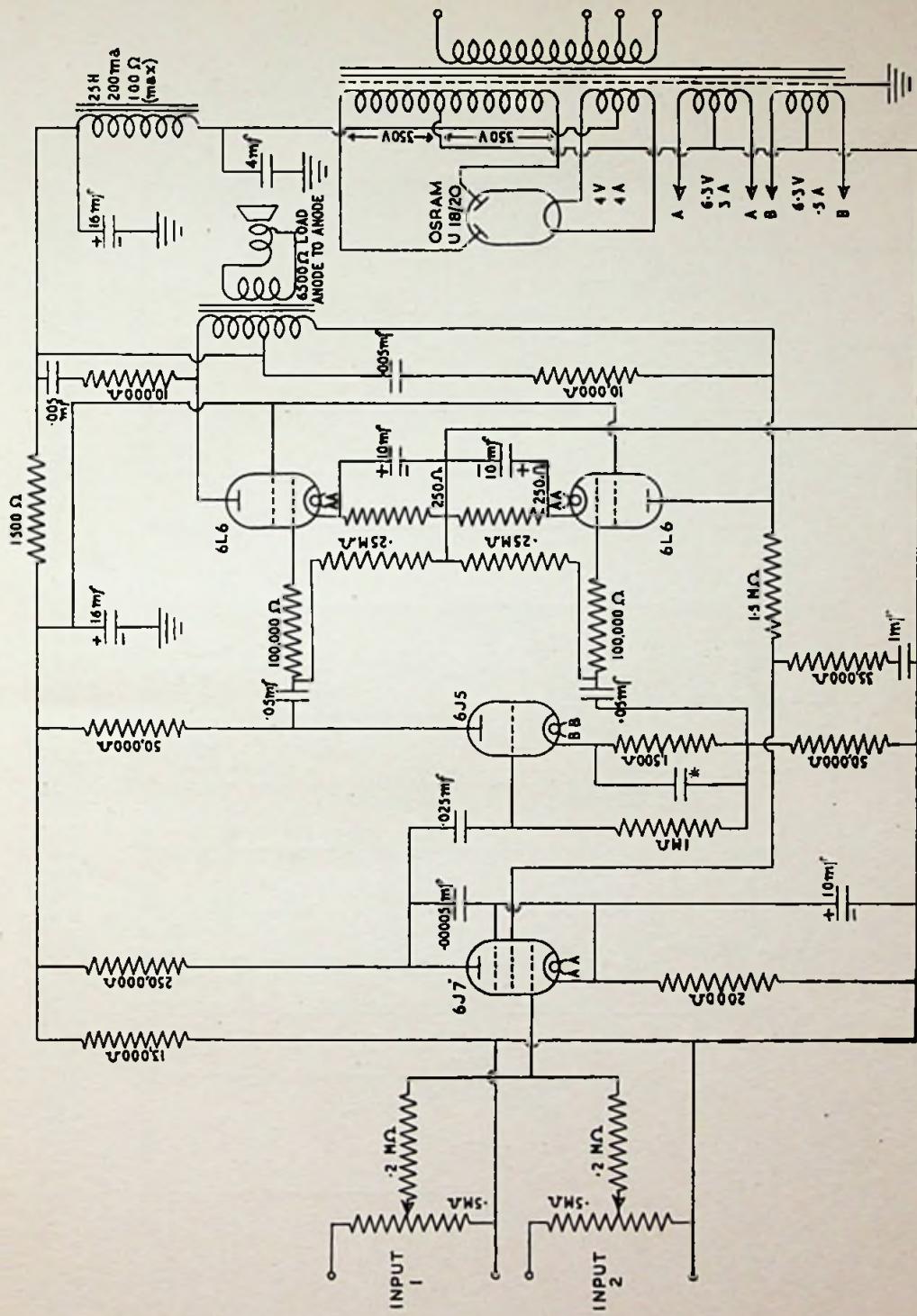
21-WATT Q.P.P. BATTERY AMPLIFIER

A 21-watt battery amplifier with a simple two-input mixer. The resistance marked \dagger will normally require selection to ensure that the first amplifier works under optimum conditions. The negative feed-back resistances marked * are best determined by trial, 100,000 ohms being a convenient starting-point. A variable top cut control and a fixed bass cut device are fitted.

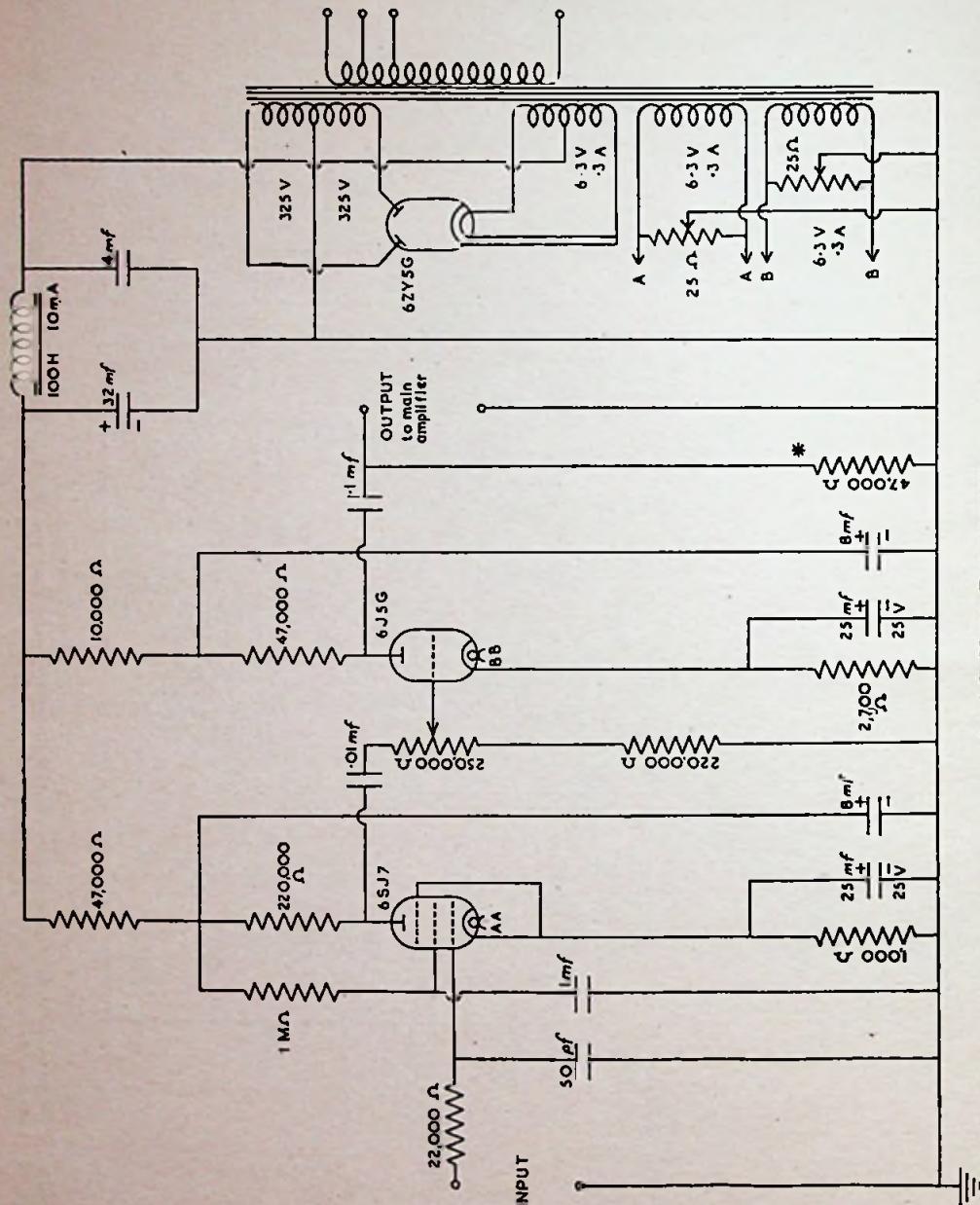


10-WATT A.C. MAINS AMPLIFIER

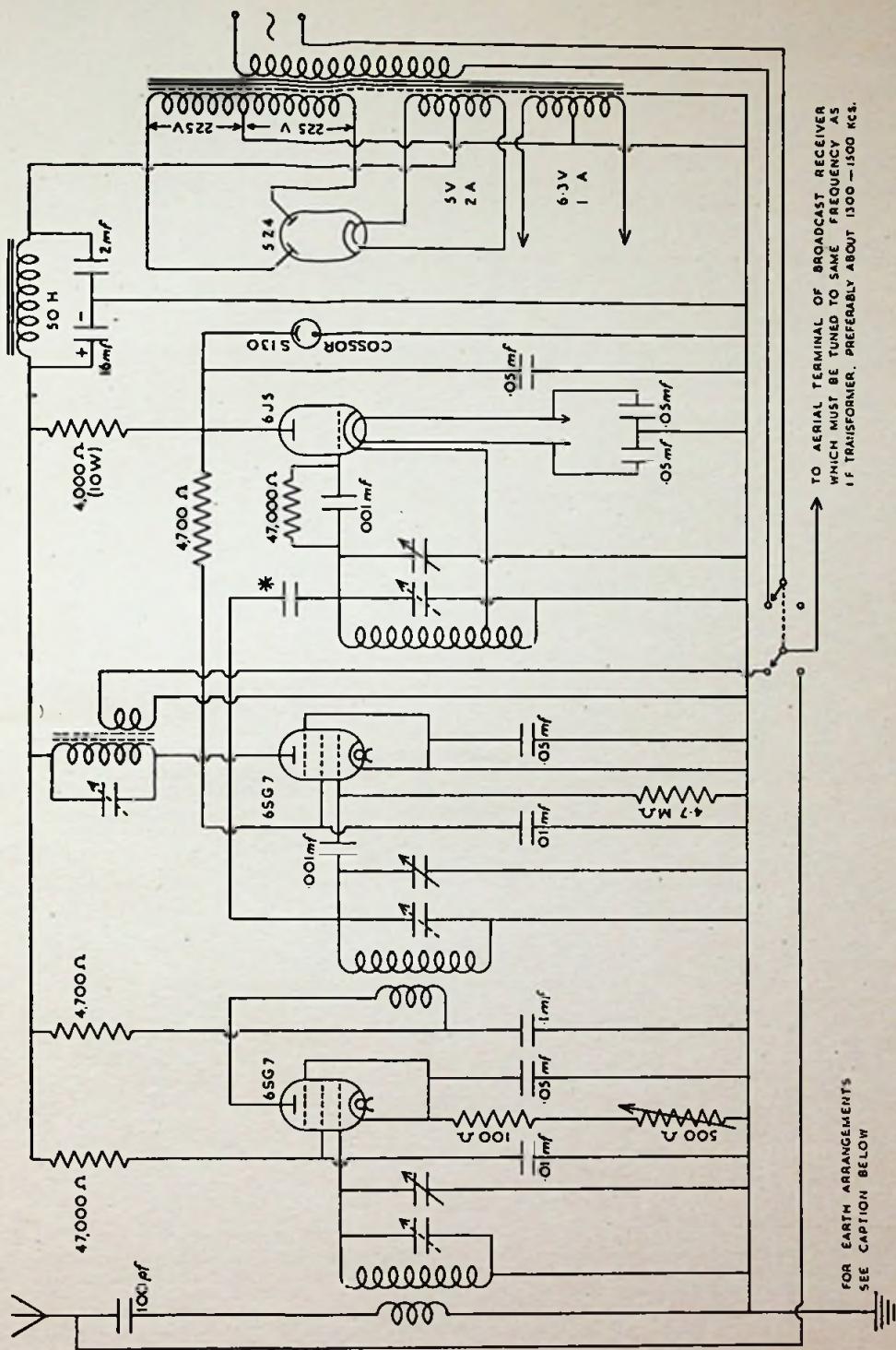
A conventional 10-watt amplifier employing negative feed-back it is essential that the two halves of the L.S. transformer primary are tightly coupled. Tone control is omitted as most users will prefer separate tone controls associated with each input.



20-WATT AMPLIFIER
An amplifier similar to that shown on page 28, but having an output of 20 watts. The condenser marked with an asterisk was found to have a marked effect on stability in the original amplifier, the value being 10 p.f.

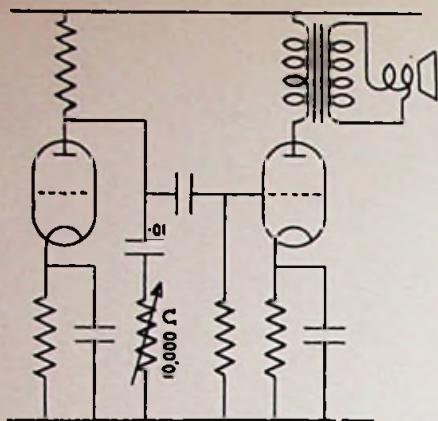


AN A.C. MAINS PREAMPLIFIER
 A typical preamplifier for use with crystal or other low output microphones. With certain types it will be necessary to connect the appropriate resistive load across the input terminals; in any case a D.C. path must be provided for the grid of the first valve. Resistance marked * should be omitted when the main amplifier already has a D.C. connection to the grid of its first valve.

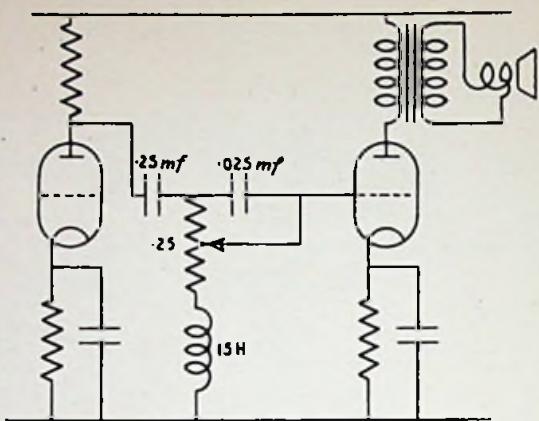


SHORTWAVE CONVERTER FOR A.C. MAINS

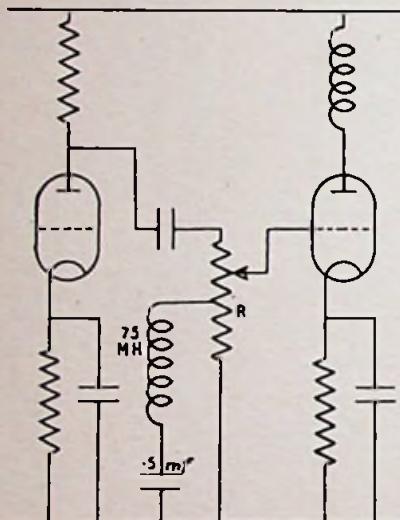
A particularly efficient shortwave converter with H.F. amplifier. The aerial coil and H.F. transformer can conveniently be ganged and the oscillator coil separately tuned or the three may be ganged together, in which case padding condensers can be introduced in the oscillator coil or coils. The condenser marked * may take the form of two wires twisted together or a small pro-set, and should be the subject of experiment to obtain maximum results. If the earth lead of the main receiver is short, earth the unit separately; if not, join the two earth terminals together and use a common earth lead. The I.F. transformer should preferably be about 1,300-1,600 kcs, and should have about two-thirds of the secondary removed.



The conventional tone-control circuit for the variable attenuation of the high frequencies.

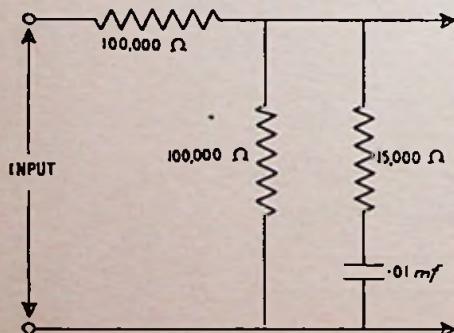


A tone-control circuit for the variable attenuation of the lower frequencies.

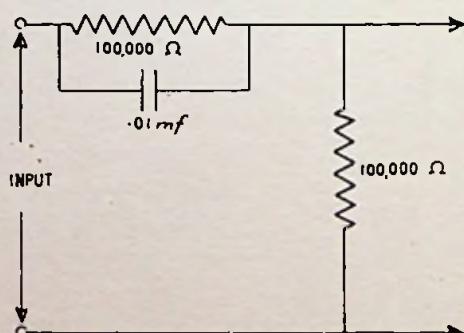


(Above) Bass boosting circuit suitable for use with an average magnetic pick-up. Resistance marked * may be variable, in which case it may be 50,000 Ω .

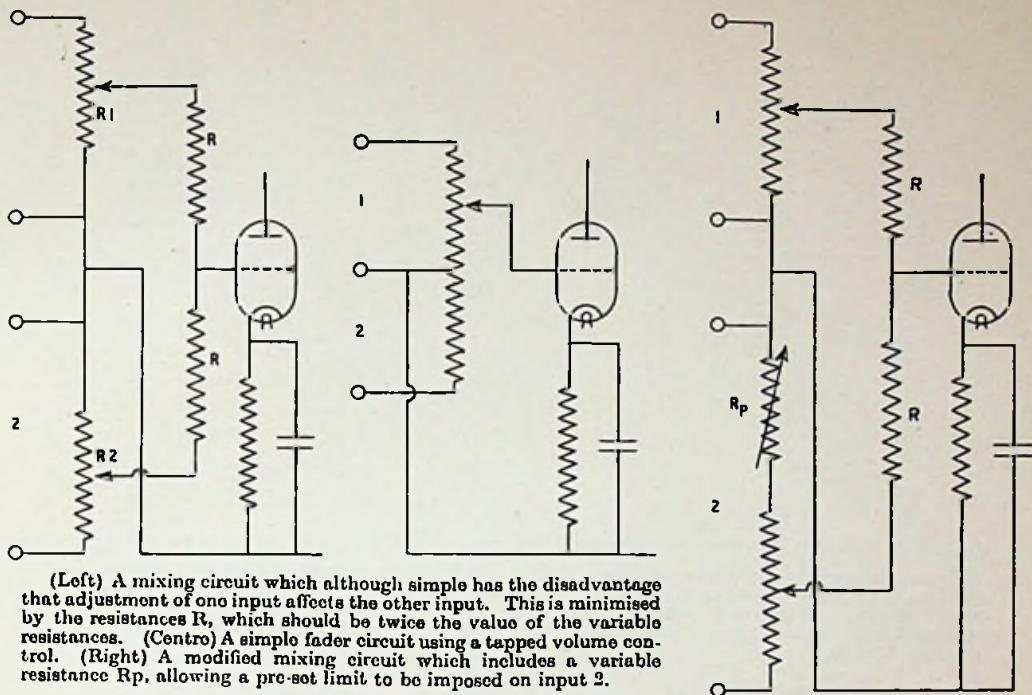
(Left) Compensated volume control avoiding the apparent serious change in frequency response when volume is varied. R should be 30,000 Ω , the upper portion being 24,000 Ω and the lower 6,000 Ω .



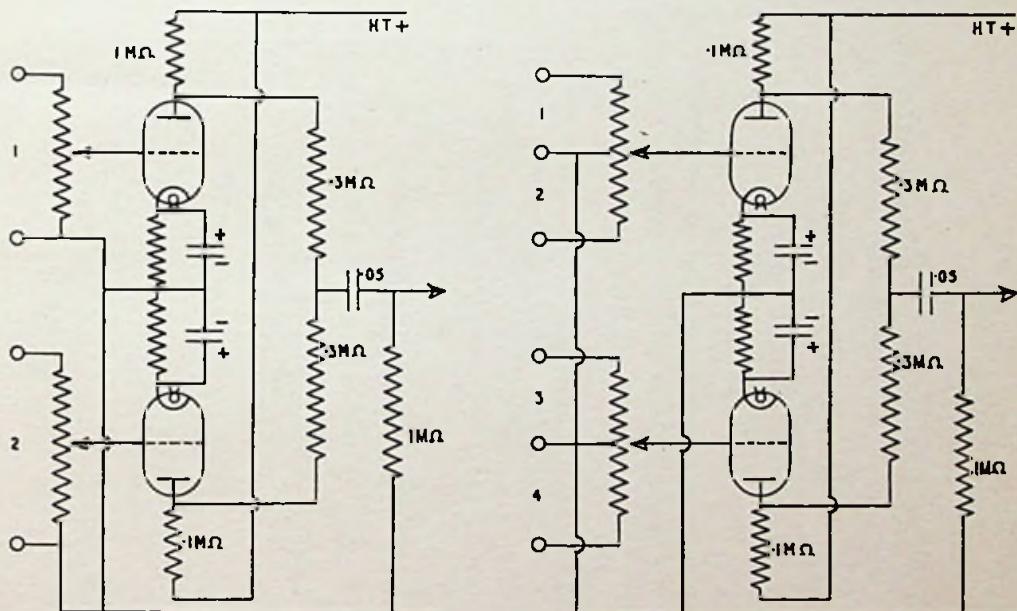
A filter giving a rapid decline in response from 50 to 1,000 cycles followed by a sensibly flat characteristic; useful when really heavy bass response is desired.



A filter giving sensibly level response up to 100 cycles followed by a rapid increase between 100 and 1,000 cycles, after which the increase tends to become progressively more gradual. Useful for boosting the middle frequencies and increasing the intelligibility of speech.

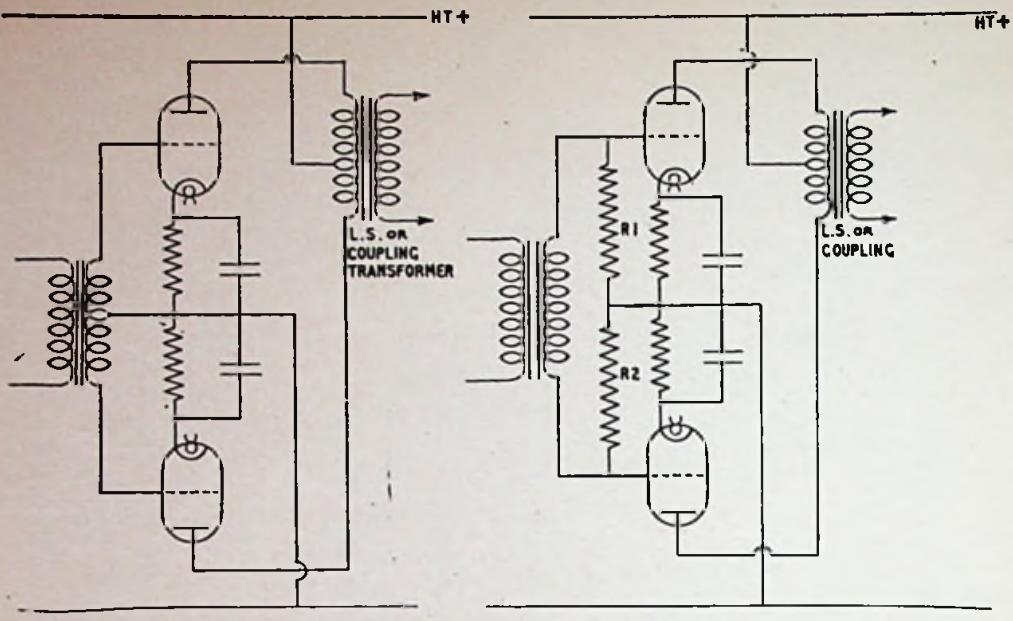


(Left) A mixing circuit which although simple has the disadvantage that adjustment of one input affects the other input. This is minimised by the resistances R , which should be twice the value of the variable resistances. (Centro) A simple fader circuit using a tapped volume control. (Right) A modified mixing circuit which includes a variable resistance R_p , allowing a pre-set limit to be imposed on input 2.



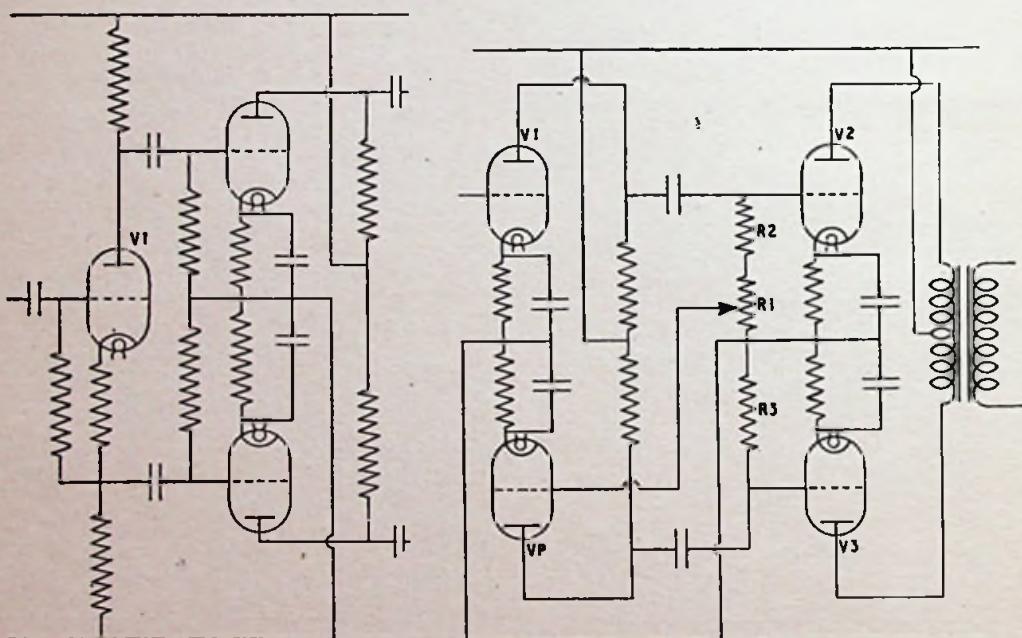
A twin-valve mixing circuit giving completely independent adjustment of both inputs. Values shown are suggested for a fairly high impedance triode.

A similar arrangement to that shown (left), but giving combined fading and mixing for two pairs of inputs.



The simplest method of phase splitting, using an input transformer with centre-tapped secondary. Separate bias resistances are shown, but a common bias resistance can be used if desired.

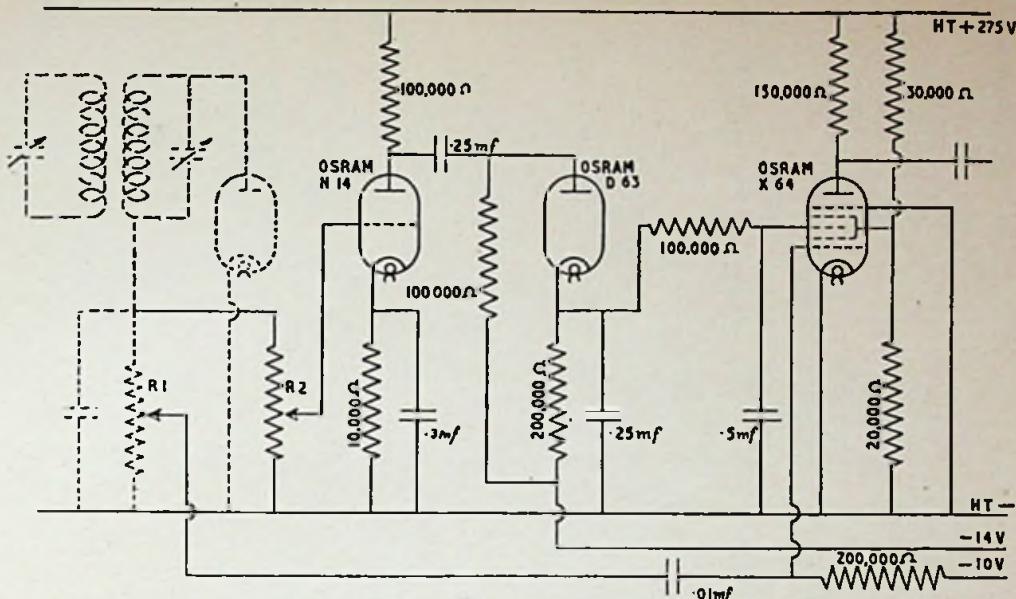
A variation of the circuit (left) sometimes advocated when a centre-tapped transformer is not available; it is a bad arrangement, since the value of $R_1 + R_2$ divided by the square of the transformer ratio appears as a reflected load in the anode circuit of the preceding valve.



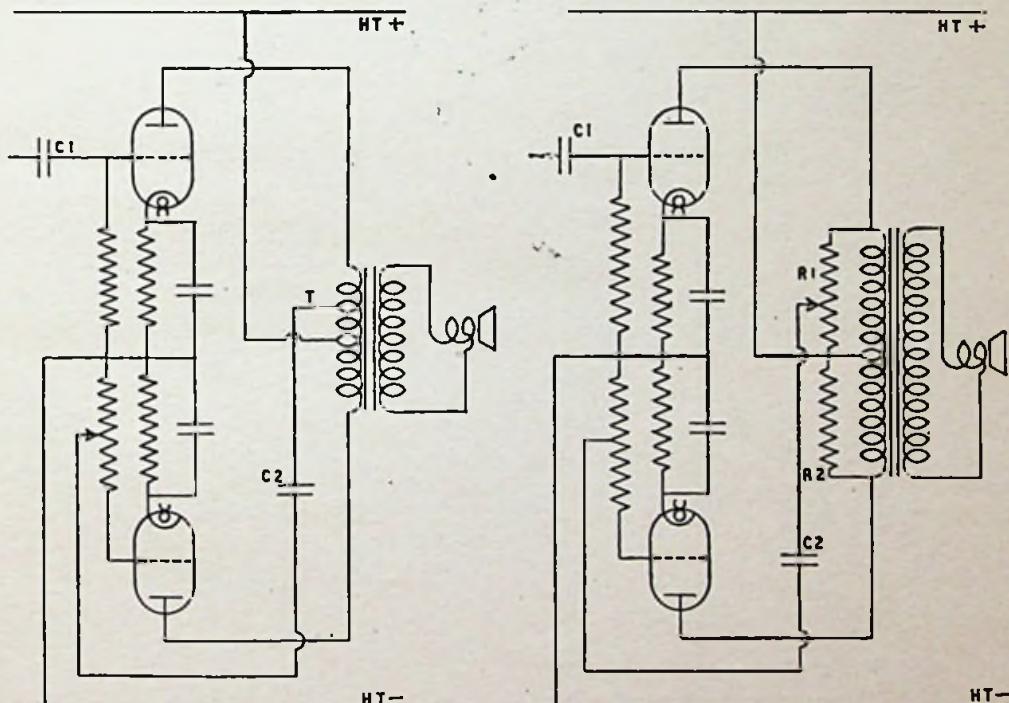
Probably the most popular phase-splitting arrangement; the anode load of V_1 is halved, one half being placed between anode and H.T. +, the other between cathode and H.T. -.

A very useful phase-splitting arrangement which offers a simple means of achieving balance. $R_1 + R_2$ should equal R_3 , R_1 being just large enough to give the required input to the paraphase valve V_p . This valve makes no contribution to the stage gain of the amplifier.

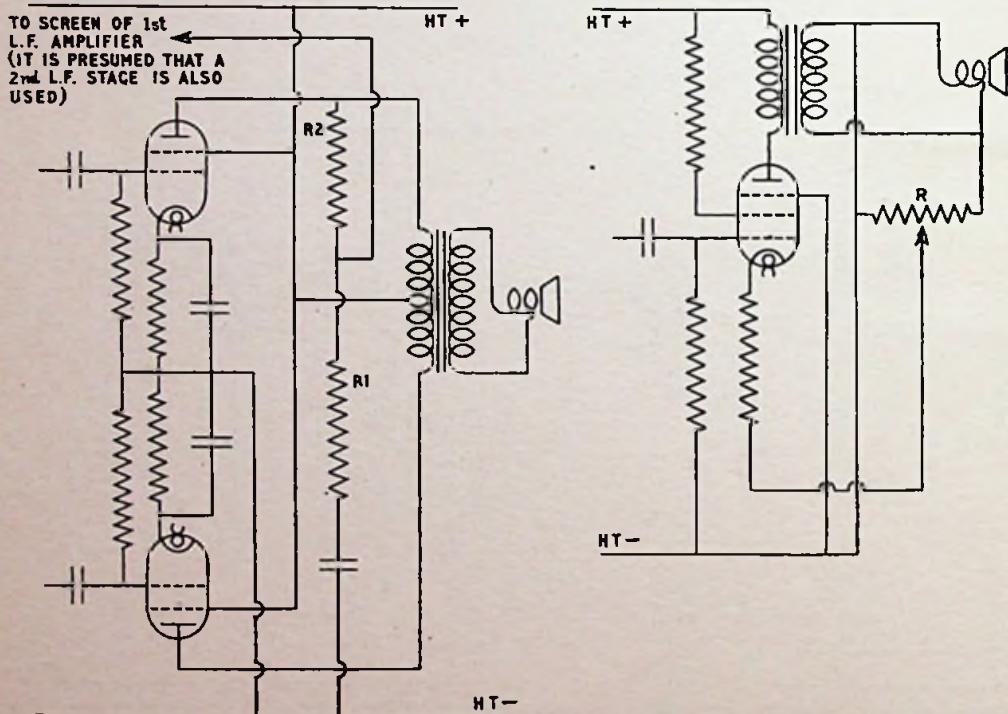
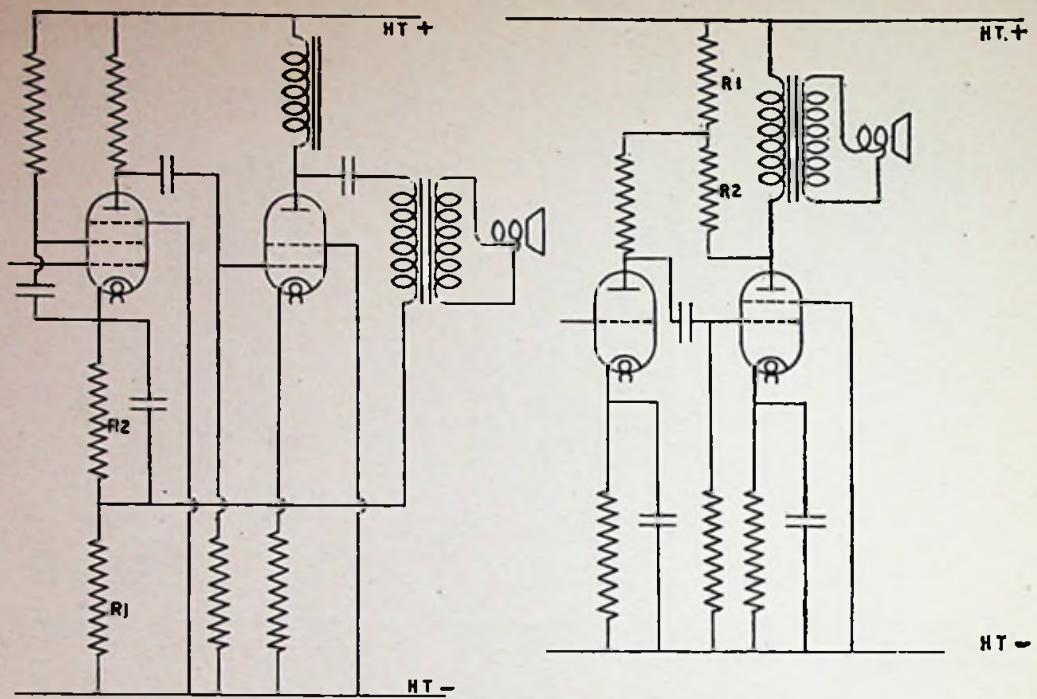
PUSH-PULL, PHASE SPLITTING



A typical volume expansion circuit, the shaded portion represents a normal diode detector. Considerable difficulty is usually experienced in getting volume expansion to function correctly, and some experiment is invariably necessary.

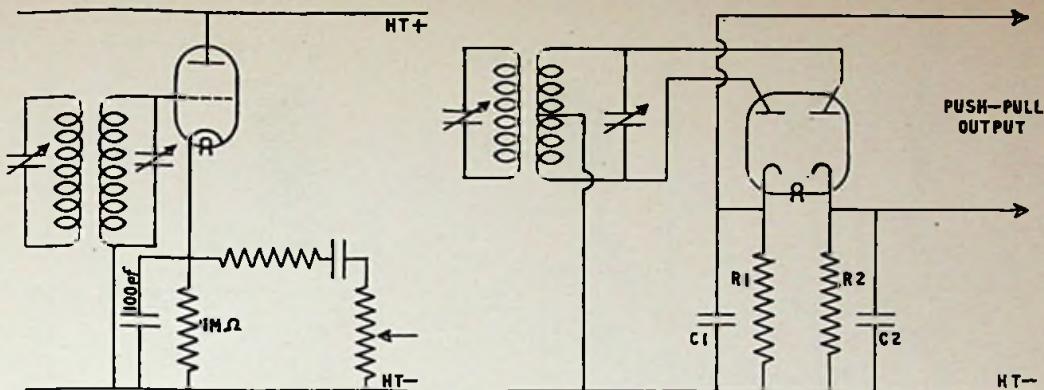


Two-phase splitting arrangements which have serious shortcomings but are very convenient when converting a single output stage to push-pull. The circuit (left) is preferable, but requires a spare tap T on the output transformer primary. The circuit (right) does not require a tap. R1 should be equal in value to R2, and together their resistance should be about ten times the primary impedance. In both circuits C1 is the coupling condenser and C2 the H.T. isolating condenser, and can be about .1 mf and 1 mf respectively.



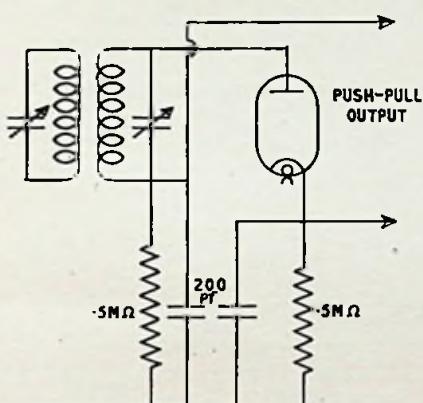
In these typical negative feed-back circuits, where resistances are so marked, feed-back increases as R_1 increases in relation to R_2 . In the circuit (top left) $R_1 + R_2$ must equal the correct bias resistance value for the valve. R in the arrangement (bottom right) should be about five times the speech-coil impedance, which must be reasonably low; if positive instead of negative feed-back is obtained, reverse connections between speech coil and resistance R .

TYPICAL NEGATIVE FEED-BACK CIRCUITS

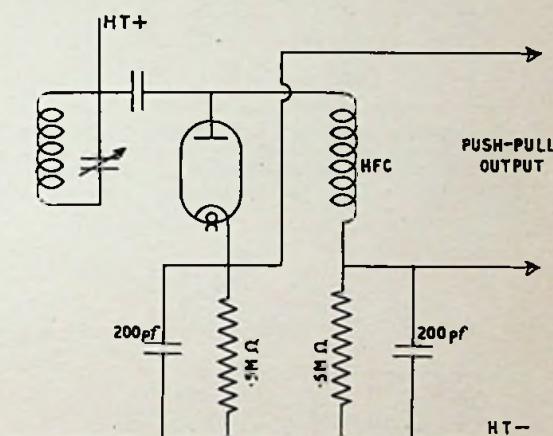


The infinite impedance detector. The volume control is usually about 5 megohm; the unmarked resistance is intended as a filter and is usually about 10,000 ohms; a suitable value for the blocking condenser is 1 mf.

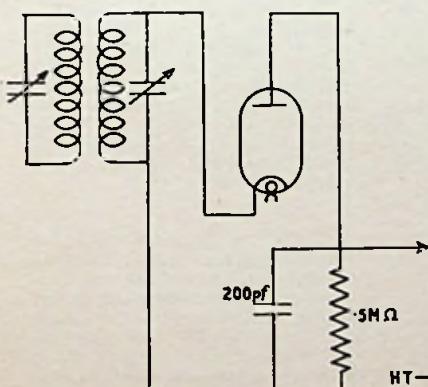
A double diodo detector arranged to deliver a balanced output, for feeding a push-pull amplifier. R₁, C₁ and R₂, C₂ form the diode loads and may be of conventional values.



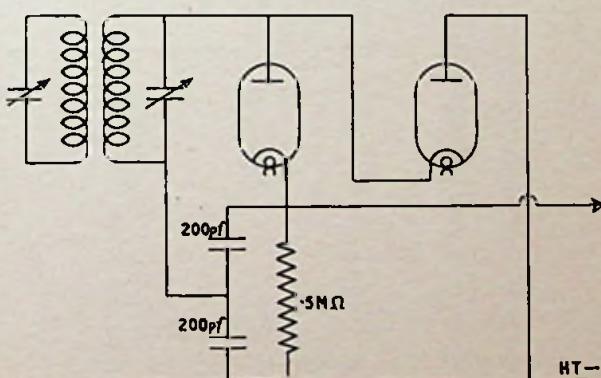
A single diode arranged to deliver a balanced output. It is, however, difficult to reduce mains hum to acceptable limits when using valves of normal construction.



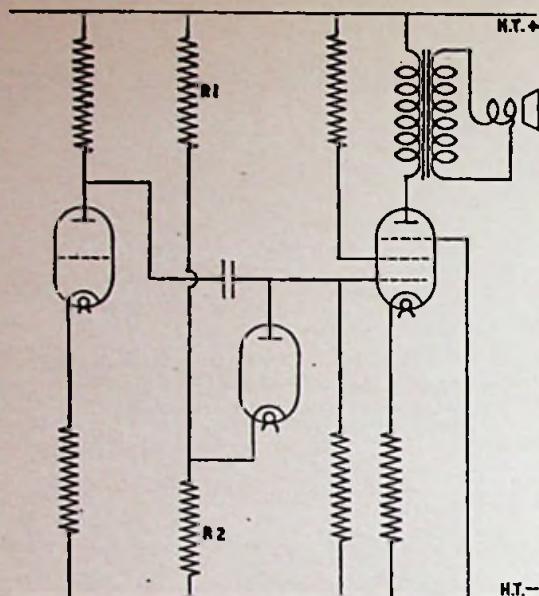
An unusual arrangement giving a balanced output from a single diode for use when the tuned circuit must be isolated in so far as D.C. potential is concerned.



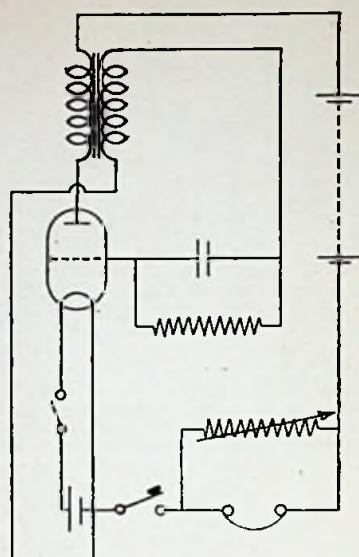
A diode detector so arranged that H.F. is virtually absent across the diode load.



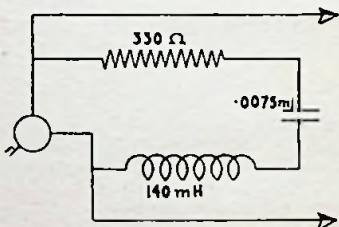
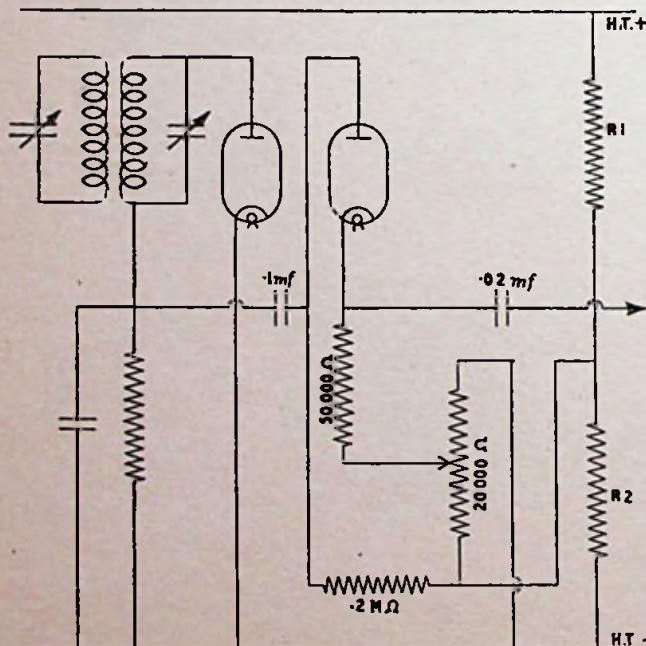
Two diodes arranged for voltage doubling which is sometimes useful in control circuits of various kinds.



A simple noise-limiter circuit which will mitigate the effect of noise of short duration, the amplitude of which is greater than that of the desired signal. As R2 is increased in respect to R1, the voltage at which the diode commences to cut is increased.



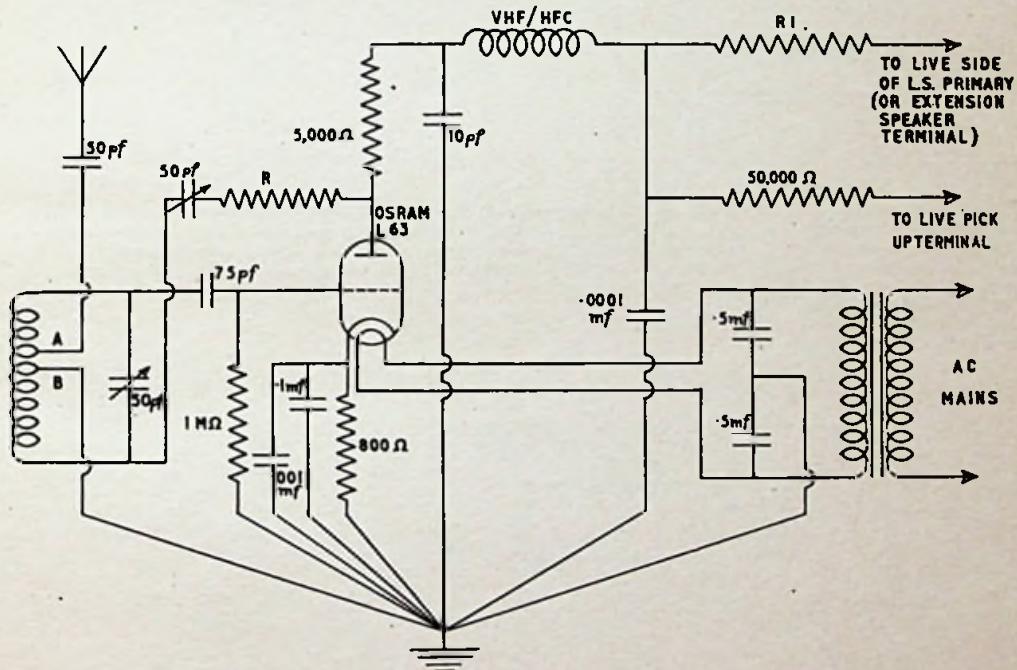
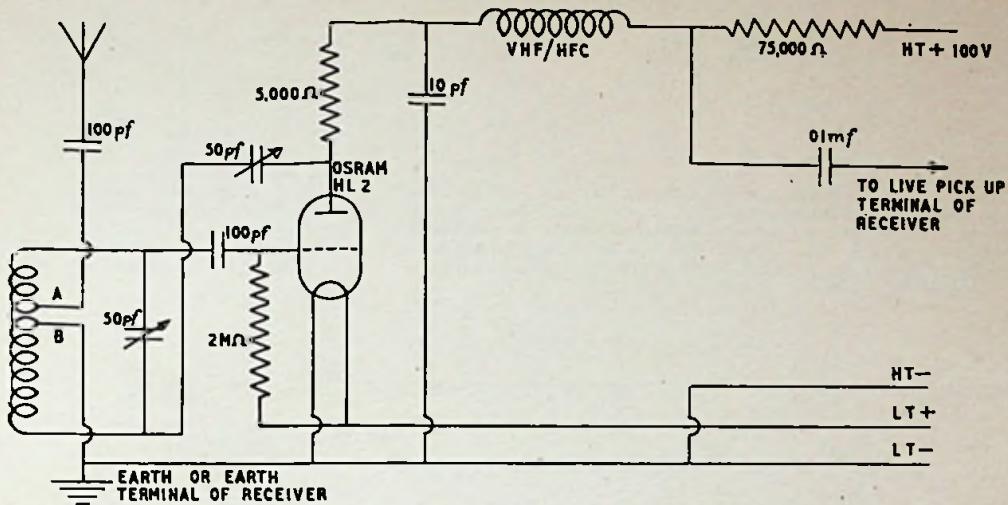
A simple oscillator suitable for Morse-code practice. With the average triode, 12-20 volts H.T. is usually sufficient. Suggested values for grid leak and condenser are .25 megohm and .01 mfd.



(Above) A tuned scratch filter suitable for use with the average magnetic pick-up. The resistance should have a value of 330 ohms *including* the D.C. resistance of the actual pick-up. If necessary, the condenser value shown can be obtained by connecting the following values in parallel: .005 mfd + .002 mfd + .0005 mfd.

(Left) An improved form of noise limiter, immediately following the detector. The variable resistance controls the amplitude at which limiting commences: R1 can have a value of 100,000 ohms for each 100 volts of H.T. supply; R2 is selected to give the required control.

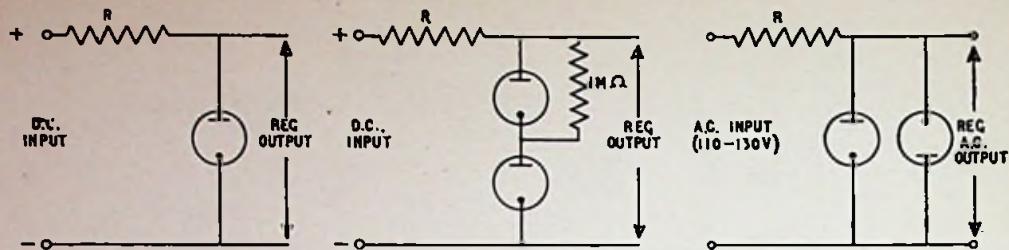
NOISE LIMITING AND SCRATCH FILTER CIRCUITS



TELEVISION SOUND ADAPTERS

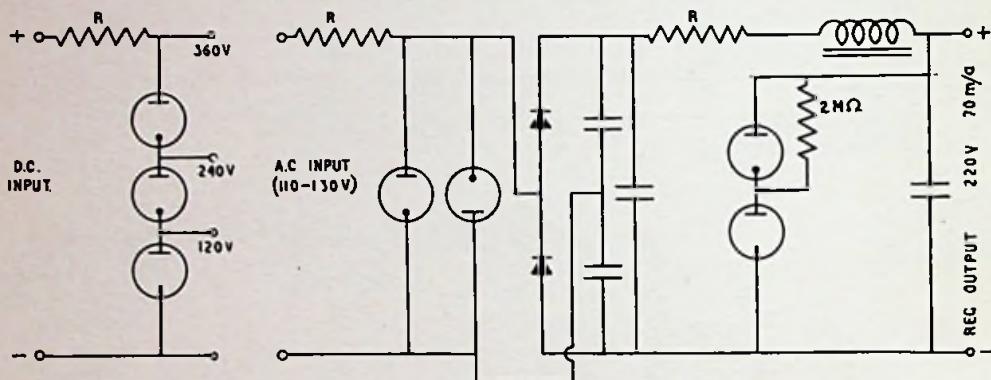
These adapters enable the sound which accompanies television broadcasts to be reproduced on receivers not provided with this facility. The arrangement (top) is for use with a battery receiver, but can also be used with a mains receiver if an accumulator is employed for heating the filament. The coil consists of five turns of 20 S.W.G. wire spaced equal to wire diameter and wound on a $\frac{1}{4}$ -inch diameter former, which should be removed, leaving the wire self-supporting. The top tap is two turns from the top and the lower tap between a half and one turn lower.

The A.C. mains version uses the same coil as the battery version and employs a transformer for heater supply: the secondary voltage should be 0.3 volts with the valve shown, or 4 volts if more convenient, when an Osram MHL4 may be employed. RI should have such a value that the anode voltage is approximately 160; R is intended to smooth reaction, and its value should be found by trial.



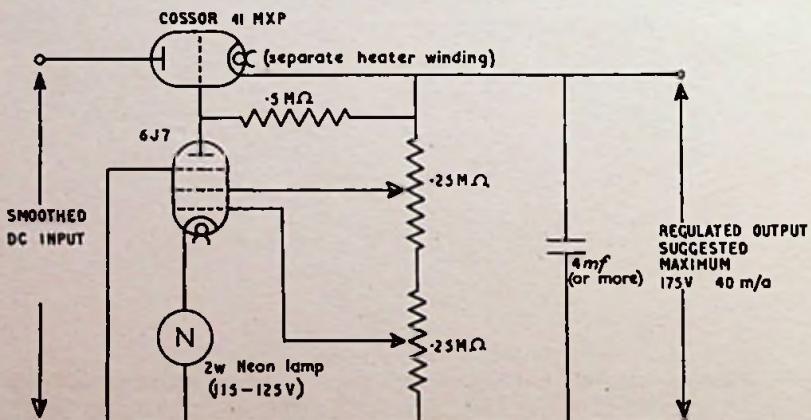
(Left) A simple circuit giving a sensibly constant voltage output irrespective of the current drawn within stated limits. Stabiliser tubes are made by various manufacturers, who publish curves from which the value of R may be determined. The controlled output voltage is usually 120/130 volts. (Right) In this arrangement two tubes are used to double the voltage output.

Two voltage stabiliser tubes in opposite sense, arranged to give a regulated A.C. output. The input voltage shown, 110/130 volts, is R.M.S. voltage, not peak voltage.



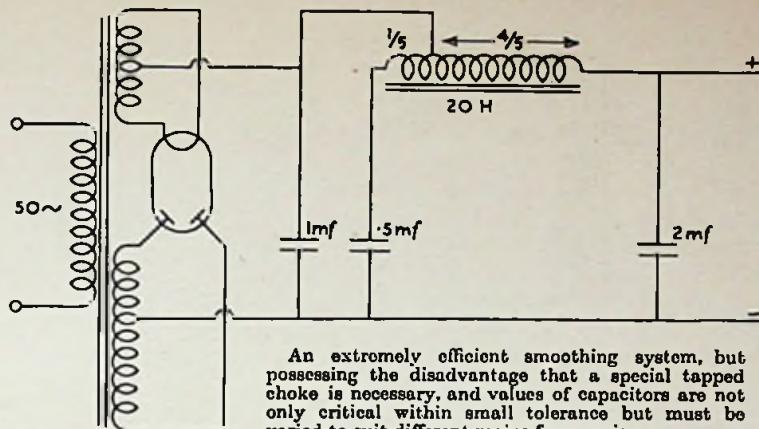
Stabiliser tubes arranged as a voltage divider. This circuit is only suitable where current drawn from the 120-V and 240-V tappings does not exceed about 2 mA each.

An elaborate arrangement in which voltage regulation is achieved in both A.C. and D.C. portions of the circuit. A voltage doubler is employed to compensate for the relatively low-voltage A.C. input. When the value of R in the D.C. circuit is determined, the D.C. resistance of the choke must be considered to be part of it.

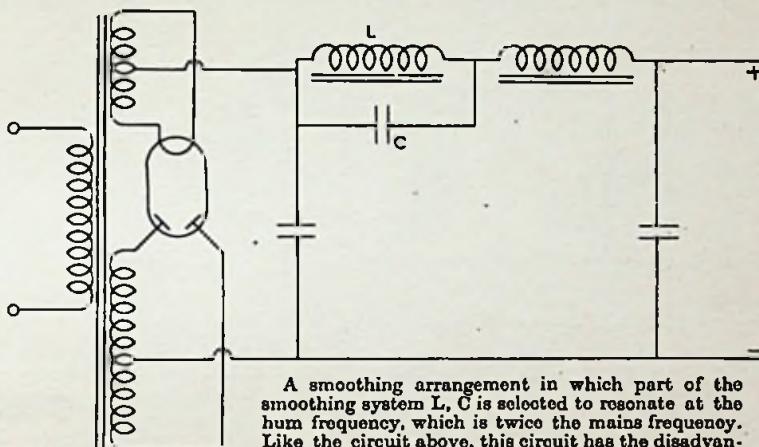


A valve voltage regulator circuit suitable for laboratory use which gives a remarkably constant output voltage. In exacting conditions where the output voltage must be independent of ambient temperature, the neon lamp must be replaced by some source of D.C. bias such as the 24-volt accumulator.

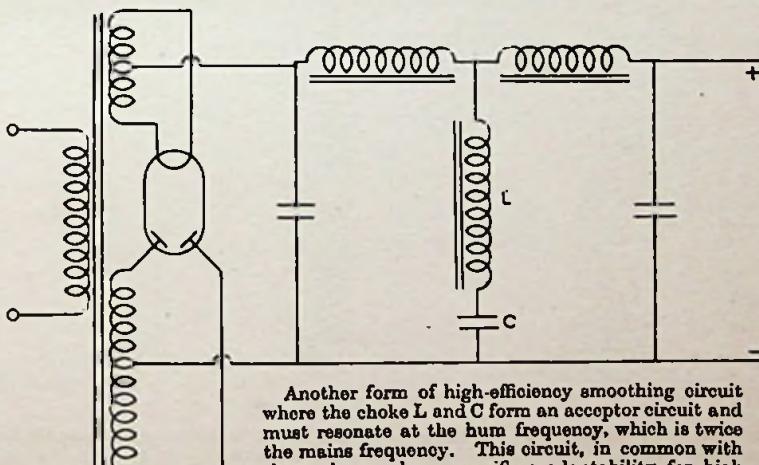
VOLTAGE STABILISATION



An extremely efficient smoothing system, but possessing the disadvantage that a special tapped choke is necessary, and values of capacitors are not only critical within small tolerance but must be varied to suit different mains frequencies.

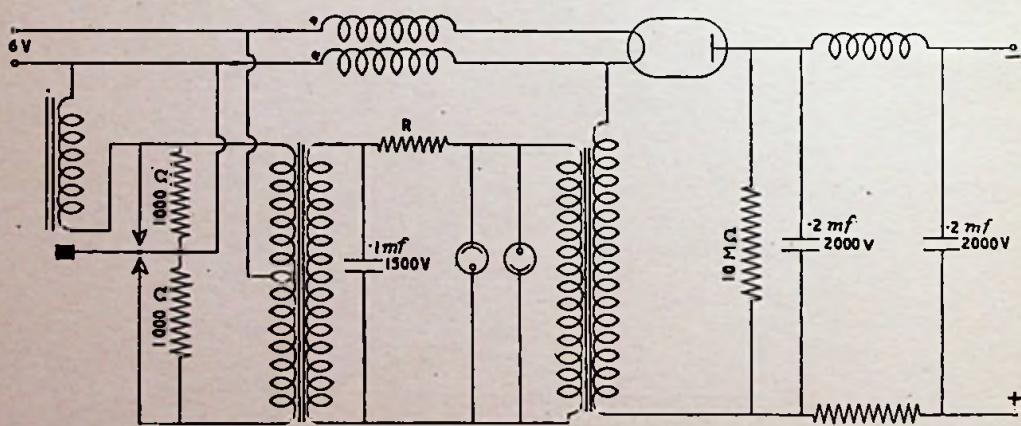
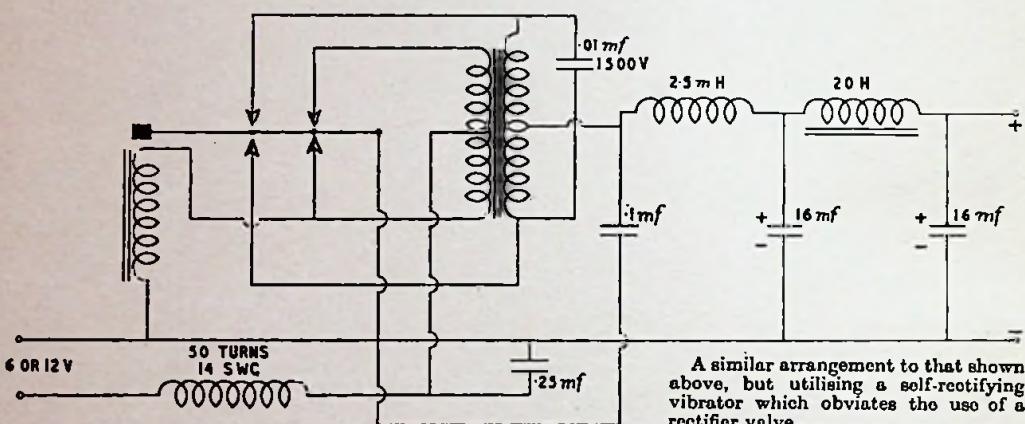
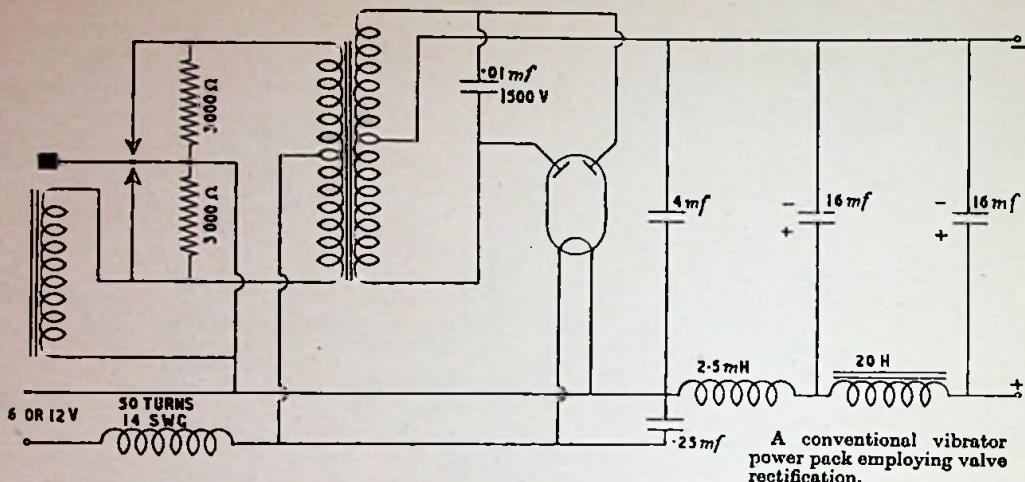


A smoothing arrangement in which part of the smoothing system L , C is selected to resonate at the hum frequency, which is twice the mains frequency. Like the circuit above, this circuit has the disadvantage that values must be selected for a particular mains frequency.



Another form of high-efficiency smoothing circuit where the choke L and C form an acceptor circuit and must resonate at the hum frequency, which is twice the mains frequency. This circuit, in common with those shown above, sacrifices adaptability for high efficiency.

UNUSUAL SMOOTHING CIRCUITS



An interesting vibrator power pack used by the author for the E.H.T. supply of a portable oscilloscope which worked from a 6-volt car battery. The author has used a similar circuit to obtain the E.H.T. supply for a television receiver converted for use on D.C. mains. H.F. chokes marked * should be inter-wound; R should be adjusted so that the neon stabiliser tubes only just glow. The $10M\Omega$ resistance must be made up of several resistances in series, e.g. five resistances of $2M\Omega$ each.

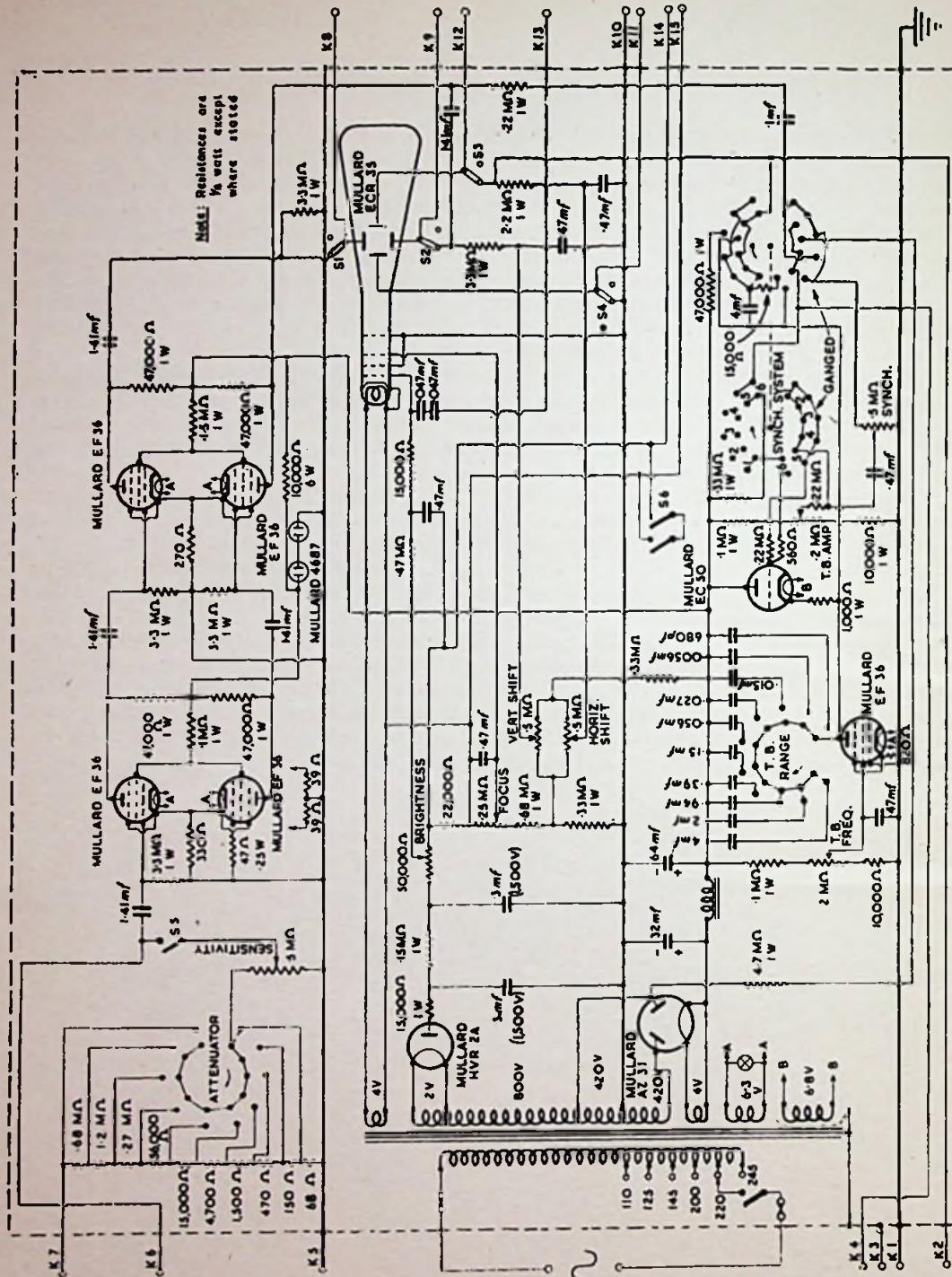
TEST GEAR AND SERVICING EQUIPMENT SECTION

PUBLISHED test gear circuits are very often idealistic in character but presented without component values; this is usually due to the fact that the individual or company responsible for the design feels disinclined to give away the result of weeks or months of patient development work. With the exception of the oscilloscope on page 44, which is a recently designed commercial instrument, the circuits which form the subject of this section do not pretend to be idealistic, but have the compensating advantage that all component values are given; furthermore, they can be relied upon to function properly, since they are precise copies of miscellaneous pieces of test gear built by the author and used for varying numbers of years with the object of reducing wear and tear on precision laboratory instruments when the use of such accurate instruments is not essential.

Those who contemplate constructing one or more pieces of test gear must realise that a considerable amount of thought and care must be expended if such equipment is to be even moderately accurate. Components must be selected with great care, mechanical construction must be above criticism, while certain instruments such as signal generators and valve voltmeters must be efficiently screened.

Any component which controls or affects the calibration of a piece of test equipment must be a very great improvement upon the type of component used in the construction of the ordinary radio receiver. If this remark applies to any one type of component more than another, it is to the variable condenser, mass-produced examples of which cannot be relied upon to repeat the same capacity to within the required limits each time the moving section is rotated to the same position.

When constructing test gear, it must always be borne in mind that the assembly as a whole must possess sufficient mechanical rigidity to prevent calibration from being affected by normal rough handling or the position in which the equipment rests. In conclusion, let it be said that a piece of inaccurate or unreliable test gear is a most dangerous thing to have on the service bench or in the laboratory, and one of the greatest potential sources of wasted time, patience, and energy.



A TYPICAL OSCILLOSCOPE
(See page 45 for description)

A. Typical Oscilloscope

The circuit diagram on the facing page shows a typical oscilloscope, the Mullard E800, which is included as representative of present-day oscilloscope practice. It is neither economical nor practicable to attempt the individual construction of such a comprehensive instrument, but by careful study the various functions can be dissected and the relative parts of the circuit used to form the basis of any workshop instrument that may be contemplated, or as a hook-up with an existing oscilloscope that lacks some desired facility. Voltages of the order of 1,500 volts are normal with instruments of this type, and in the case of the example illustrated, two external sockets, K14 and K15, are at a potential of 1,250 volts with respect to earth.

Brief characteristics are : Time-base frequency range 25—16,000 cps. Vertical deflection sensitivity 7V.A.C. (R.M.S.) per cm. or 2IV.D.C. per cm. Horizontal deflection sensitivity 1.4V. A.C. (R.M.S.) per cm. or 42V. D.C. per cm. The voltage amplifier has a gain of 7,000 approx. with a frequency response from 1—40,000 cps. flat within 2 db. The synchronism system switch permits a choice of six systems ; positions 1, 2, 3 and 5 use the internal time-base. Position 4 permits the use of an external time-base source, while position 6 is single stroke. Synchronism for positions 1, 2, 3 and 5 is derived as follows : (1) From vertical deflection amplifier. (2) From external source applied between K3, K4. (3) From 50-cycle mains. (5) By means of mechanical contact breaker S6 ; the gas discharge triodo fires at "break."

There are fifteen external connections and four external links, which are intended for use in the following manner and give the facilities implied:

K1, K2. Time-base voltage outlet for external use ; available when synchronising system switch is in position 4.

K3, K4. Same use as K1, K2, but for use when synchronising system switch is in position 4.

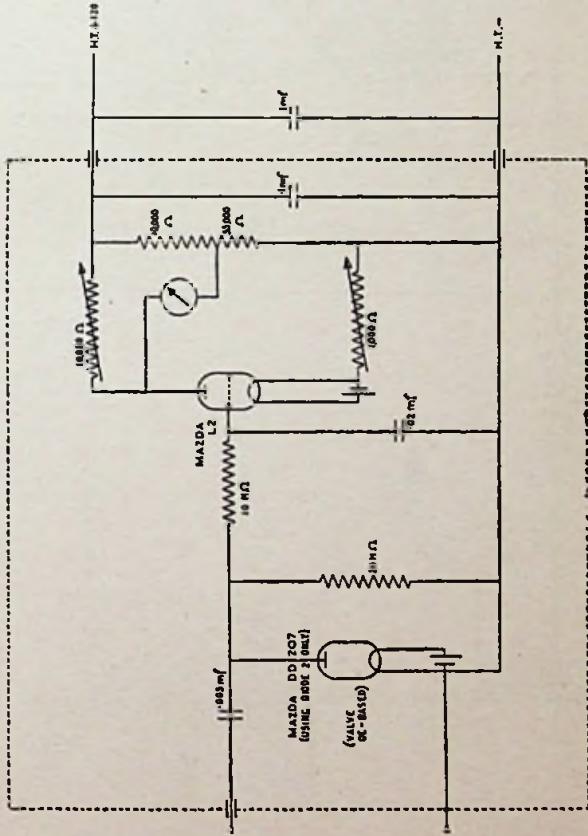
K5, K7. Inlet for voltage to be observed ; when S5 is opened for high impedance input, K5 and K6 are used instead of K5 and K7.

K8, K9. Give direct access to vertical deflection plates, which can be freed from internal connection by opening the links S3, S4.

K11, K12. Give direct access to horizontal deflection plates, which can be freed from internal connection by opening the links S3, S4.

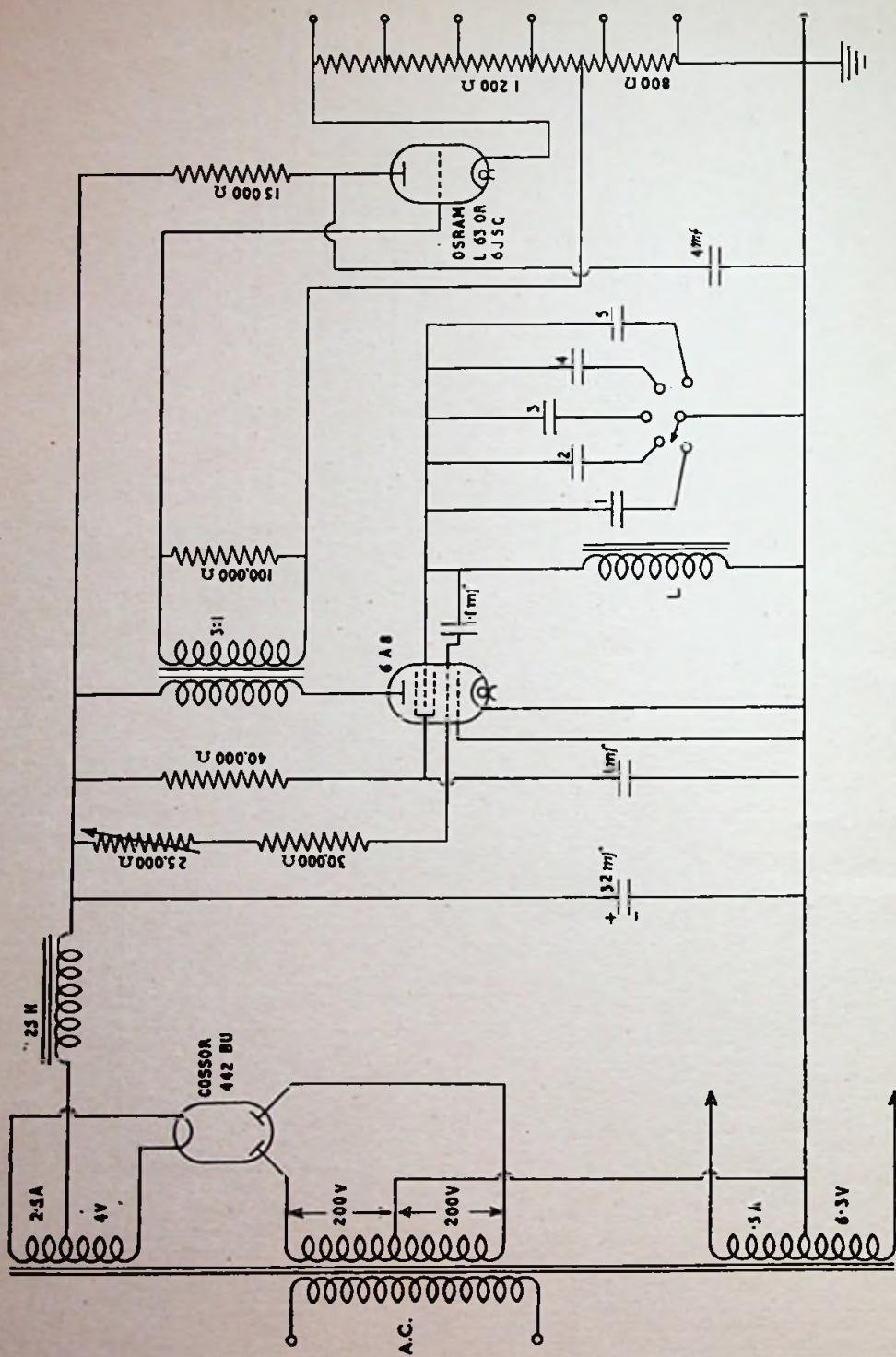
K10, K13. Inlet for grid modulation voltage.

K14, K15. Are for beam suppression by an external source ; the beam is suppressed when K14, K15 are short-circuited.



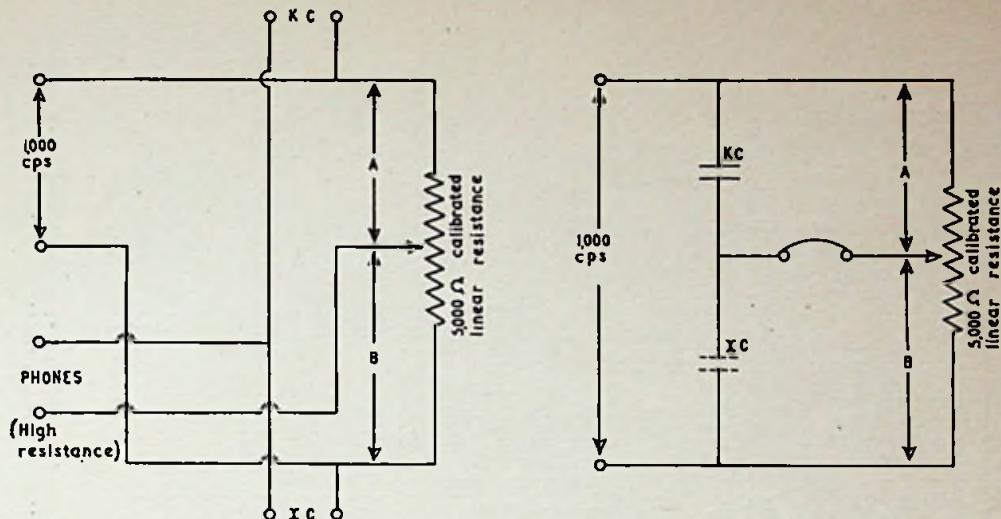
VALVE VOLTMETER

A simple, but highly efficient diode valve voltmeter with triode amplifier. Note that a separate accumulator must be used for each filament and that they must be housed within the screen or in a separate screened compartment without exposed connecting wires. Resistance values shown are satisfactory when using a standard 250- μ A meter, which will give full scale deflection with an input of only about 1.5 volts. If properly designed, calibration does not fail off suddenly with frequencies up to 50 Mcs. The 10,000 Ω resistance is so adjusted that the meter reads zero after the 1,000 Ω bias resistance has been set so that the triode is working at the beginning of the straight portion of its characteristic. Larger input voltages can be accommodated by shunting the meter, but a limitation is imposed when the input is large enough to overload the triode.



A GENERAL-PURPOSE AUDIO OSCILLATOR

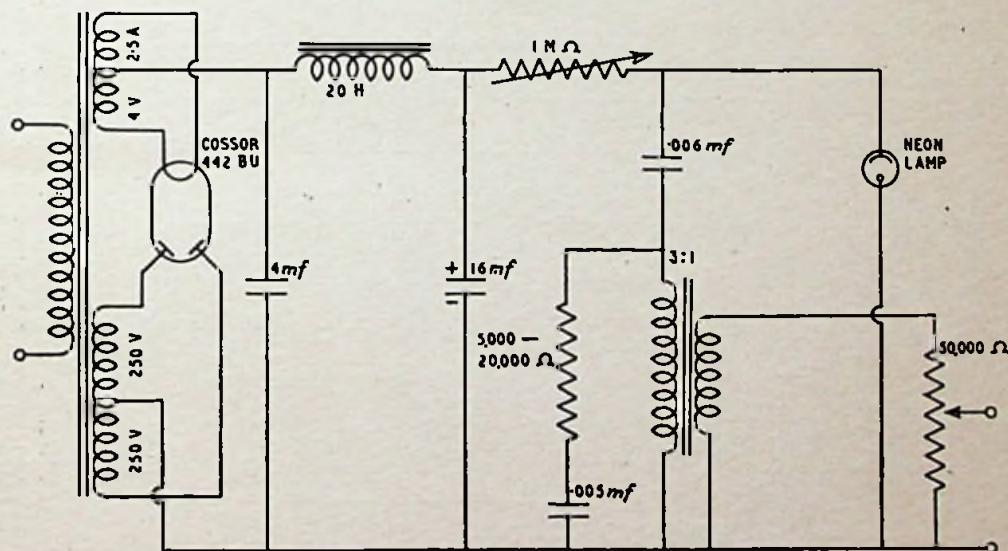
Originally constructed for general purposes, this oscillator has proved very satisfactory. The use of the L.F. transformer is open to criticism, but in practice it is quite satisfactory; an additional lead may be led out from the top end of the secondary if a high impedance output is desired. The L.F. transformer can, of course, be replaced by resistance capacity coupling, but output is thereby reduced. The output frequency is determined by the value of L and the values of the various condensers; for formula see page 82.



A simple capacity bridge, the accuracy of which is dependent on the accuracy of known condensers available and the calibration of a variable resistance which must be non-inductive. The diagram (left) is a practical wiring circuit, that on the right being a schematic arrangement to illustrate the following instructions. To find the capacity of an unknown condenser XC , connect a condenser of probable similar value at KC and adjust resistance until the injected 1,000-cycle note disappears or reaches a definite minimum, when

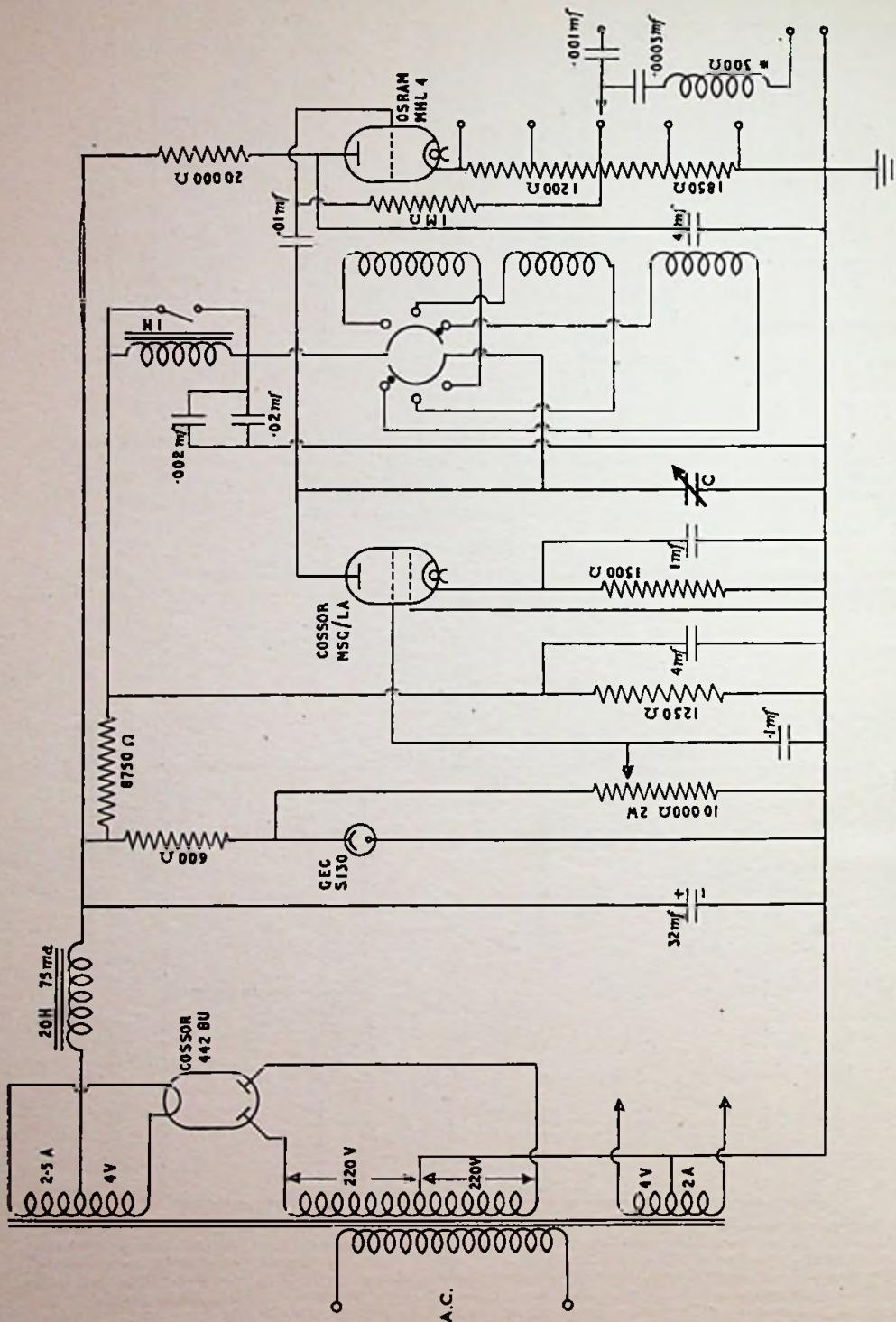
$$XC = \frac{R_A}{R_B} KC$$

when XC equals the unknown, and KC the known capacity and R_A is the portion of the resistance across KC and R_B is the portion across XC . If R_A is greatly dissimilar to R_B , greater accuracy can be obtained by selecting a condenser for KC which gives a better ratio.



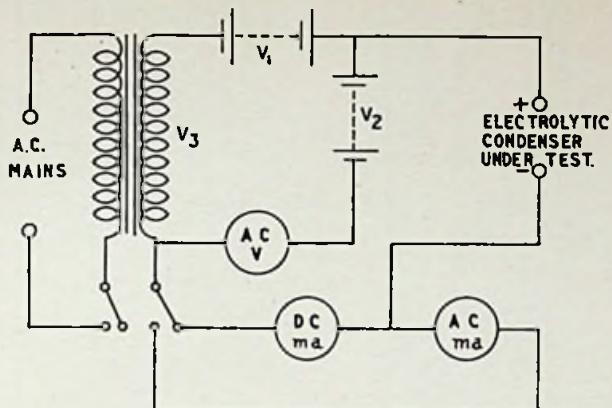
A very simple form of audio oscillator, useful where simplicity and economy are necessary and a bad waveform can be tolerated. An ordinary L.F. transformer is used, but connected to give a step-down ratio to increase output current at the expense of voltage; where higher voltage is desired at the expense of current the transformer can be reversed. A resistance and condenser in series are shunted across the transformer to limit changes of output at different frequencies. The presence of the transformer is not essential; it limits the frequency range and prevents ready calculation of frequency, but improves the waveform for general purposes.

CAPACITY BRIDGE AND SIMPLE NEON AUDIO OSCILLATOR



A DYNATRON SIGNAL GENERATOR

The signal generator shown above is not proof against criticism, but has the advantages of simplicity and robustness; it incorporates readily obtainable components and has worked satisfactorily for more than ten years. The variable condenser C must be of impeccable design, and the whole unit must be built into a screening box from which the inductances must be spaced by a minimum of 2 inches. The inductance marked with an asterisk must consist of a suitable length of resistance wire wound on a glass tube—or this component and the .0003-mf condenser can be replaced by a standard dummy aerial. The coils must be so designed and constructed that they are adequately free from temperature and other changes. The output is modulated or not at will, the modulation frequency being approximately 1,000 cps.



This electrolytic-test set will give leakage current and approximate capacity under working conditions. With switch in position shown, leakage current may be read directly from D.C. milliammeter; V1 should equal D.C. working-voltage of condenser (V2 must exactly equal V1).

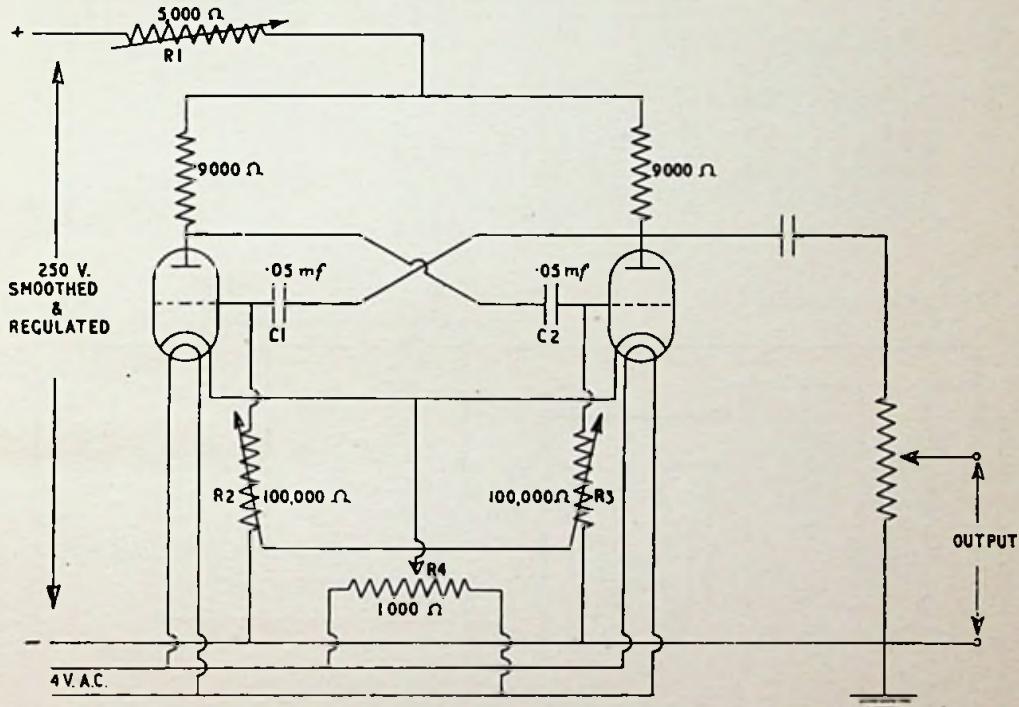
To measure capacity, throw switch and read A.C. volts and current.

$$C = \frac{159 \times I}{V \times f}$$

(C is in mf, I is in millamps; when V is in volts (R.M.S.), f is mains periodicity in cycles.)

For capacity test, V1 and V2 must be exactly equal, V3 should be about 30 per cent. of V1, and V1 + V3 should preferably be about 75 per cent. of condenser working-voltage.

Warning : Bad connections on V1 or V2 may damage the meters.



This improved form of multi-vibrator has the advantages of wide frequency range and extreme simplicity ; the multi-vibrator unfortunately, however, has a bad waveform, although with this particular arrangement some improvement can be made by careful adjustment of R1. The ganged variable resistances R2, R3 control the output frequency, R4 allows a small voltage at mains frequency to be used to lock the oscillator frequency at any desired multiple of the mains frequency.

ELECTROLYTIC CONDENSER TESTER AND MULTI-VIBRATOR

SWITCH POSITIONS

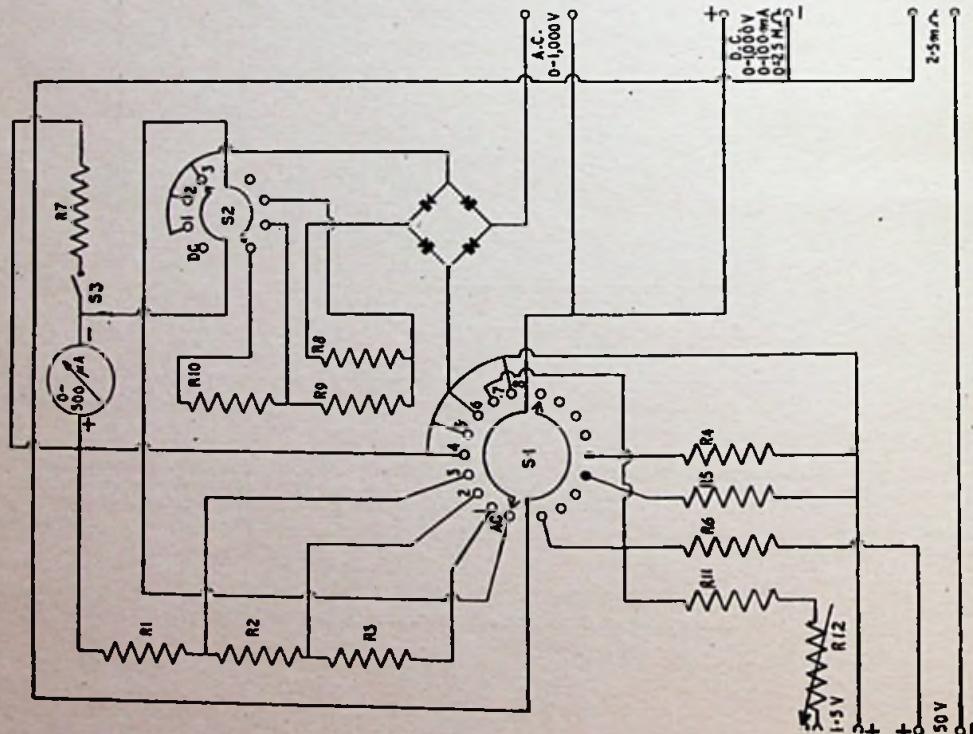
A.C. = Volts A.C.
 1 = 0-1,000 volts D.C.
 2 = 0-100 volts D.C.
 3 = 0-10 volts D.C.
 4 = 0-100 mA D.C.
 5 = 0-10 mA D.C.
 6 = 0-1 mA D.C.
 7 = 0-250,000 ohms.
 8 = 0-5000 ohms.

S2

D.C. = Ohms, D.C. volts and amps:
 1 = 0-10 volts A.C. (R.M.S.)
 2 = 0-100 volts A.C. (R.M.S.)
 3 = 0-1,000 volts A.C. (R.M.S.)

Note 1.—When switch S1 is in positions 1-8 inclusive, switch S2 must be in position D.C. When switch S1 is in position A.C., switch S2 must be in position 1, 2 or 3 as appropriate.

Note 2.—When the switch S3 is closed, the reading on the meter is halved, thus doubling all D.C. current and voltage ranges and A.C. voltage ranges except 0-10 V. A.C.



RESISTANCE VALUES

	R1	= 20,000 — R_m ohms.	1 W.
	R2	= 180,000 ohms.	1 W.
	R3	= 1,800,000 ohms.	2 W.
	R4	= $\frac{R_m}{199}$ ohms.	
	R5	= $\frac{R_m}{19}$ ohms.	
	R6	= $\frac{R_m}{85}$ ohms.	1 W.
	R7	= $\frac{R_m}{17}$ ohms.	1 W.
	R8	= 102,000 ohms.	1 W.
	R9	= 1,620,000 ohms.	2 W.
	R10	= 5,000 ohms.	1 W.
	R11	= 5,000 ohms (variable).	
	R12	= 5,000 ohms (variable).	

R_m = the resistance of the meter used, which must have a full-scale deflection of 0-600 μ A; if the internal resistance of the meter selected is 50 ohms or less, R1 can be 20,000 ohms.

Rectifier unit—Westinghouse Type 1 mA bridge instrument rectifier.

Note.—If high accuracy is required on the 0-10 V. A.C. range, R8 should be varied to suit the individual characteristics of the rectifier used.

A UNIVERSAL TEST SET

(2,000 ohms per volt)

For ranges see table above, left. R1, 2, 3, 6, 8, 9, 10 must be of the high stability 1% tolerance type. R4, 5, 7 must be wire wound. R11, 12 may be normal resistances. In most cases the values shown will not be obtainable as a single component, but must be built up with two or more as required. Switch S1 must be of a type that does not short-circuit contacts when rotated; it may be a single two-pole, nine-way wafer or two one-pole, nine-way wafers; in either case an additional wafer should be ganged on the same spindle so that contacts shown black can be doubled up to reduce contact resistance. Separate scales should be used for range groups as follows: (1, 2, 3, D.C., 1, 2, A.C.), (3, A.C.), (4, 5, 6), (7), (8). Calibration should be checked with the aid of a similar instrument of high accuracy. This instrument has its resistance range extended to 2.5 M Ω if a 50-volt battery is connected where shown; S1 must be in position 7 and readings taken from the 0.25 M Ω scale multiplied by 10. If an H.T. battery is used to supply the necessary 50 volts, the tapings should be adjusted so that a zero reading can be obtained by adjustment of R12; if it is desired to use a non-tapped battery, a variable resistance of 50,000 may have to be placed in series with the + 50V terminal.

USEFUL FORMULÆ AND DATA SECTION

The formulæ on these pages and the tables on the pages that follow are accurate to normal engineering standards and, where appropriate, they are selected for the special use of the radio engineer and are rationalised with due regard to raw materials, tolerances and ranges suitable for radio engineering.

Voltage, Current, Resistance and Power for D.C.

$$V = IR, R = \frac{V}{I}, I = \frac{V}{R}; W = IV, W = I^2R, W = \frac{V^2}{R}$$

when V = volts, I = amps., R = ohms and W = watts.

Voltage, Current, Impedance and Power for A.C.

$$V = IZ, Z = \frac{V}{I}, I = \frac{V}{Z}; W = I^2R, W = VI \cos \varphi$$

when V = volts, I = amps., Z = ohms and φ = phase angle between V and I .

Resistances in Series

$$R = R_1 + R_2 + R_3 + R_4, \text{ etc.}$$

Resistances in Parallel

$$\frac{1}{R} = \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} + \frac{1}{R_4}, \text{ etc.}$$

Specific Resistance

$$R = \rho \frac{B}{A}$$

Two Resistances in Parallel

$$R = \frac{R_1 \times R_2}{R_1 + R_2}$$

when ρ = specific resistance per centimetre cube, R = resistance in ohms, B = length in centimetres and A = area in square centimetres.

Condensers in Series

$$\frac{1}{C} = \frac{1}{C_1} + \frac{1}{C_2} + \frac{1}{C_3} + \frac{1}{C_4}, \text{ etc.}$$

Condensers in Parallel

$$C = C_1 + C_2 + C_3 + C_4, \text{ etc.}$$

Two Inductances in Parallel

$$L = \frac{L_1 \times L_2}{L_1 + L_2}$$

Two Condensers in Series

$$C = \frac{C_1 \times C_2}{C_1 + C_2}$$

Inductances in Parallel

$$\frac{1}{L} = \frac{1}{L_1} + \frac{1}{L_2} + \frac{1}{L_3} + \frac{1}{L_4}, \text{ etc.}$$

Inductances in Series

$$L = L_1 + L_2 + L_3 + L_4, \text{ etc.}$$

Note.—Above inductance formulæ assume no mutual inductance.

Inductive Reactance

$$X_L = 2\pi fL$$

when f = cps. and L = henrys.

Capacitive Reactance

$$X_C = \frac{1}{2\pi fC}$$

when f = cps. and C = farads.

Resistance and Capacity in Series

$$Z = \sqrt{R^2 + X_c^2}$$

$$V = I \sqrt{R^2 + \frac{1}{(2\pi f C)^2}}$$

$$I = \frac{V}{\sqrt{R^2 + X_c^2}}$$

when Z = impedance in ohms, R = ohms, V = volts, I = current in amps. and C = capacity in farads.

Resistance and Inductance in Parallel

$$Z = \sqrt{R^2 + X_L^2}$$

$$I = \frac{V}{\sqrt{R^2 + X_L^2}}$$

$$V = I \sqrt{R^2 + (2\pi f L)^2}$$

$$Z = \frac{RX_L}{\sqrt{R^2 + X_L^2}}$$

$$I = V \sqrt{\frac{1}{R^2} + \frac{1}{(2\pi f L)^2}}$$

$$V = I \frac{RX_L}{\sqrt{R^2 + X_L^2}}$$

when Z = impedance in ohms, R = ohms, V = volts, I = amps. and L = henrys.

Resistance, Capacity and Inductance in Series

$$V = I \sqrt{R^2 + \left(2\pi f L - \frac{1}{2\pi f C}\right)^2}$$

$$Z = \sqrt{R^2 + (X_L - X_C)^2}$$

when Z = impedance in ohms, R = ohms, V = volts, I = amps., L = henrys and C = farads.

Frequency and Wavelength

$$f = \frac{1}{2\pi \sqrt{LC}}$$

when f = cps., L = henrys and C = farads.

$$\lambda = 1,884 \sqrt{LC}$$

when L = μ H, C = μ F and λ = metres

$$f = \frac{1,000,000}{2\pi \sqrt{LC}}$$

when f = kcs., L = μ H and C = μ F.

$$\lambda = 1.884 \sqrt{LC}$$

when L = μ H, C = $\mu\mu$ F and λ = metres.

$$f = \frac{300,000}{\lambda}$$

when f = kcs., λ = metres.

$$f = \frac{300}{\lambda}$$

when f = mcs., λ = metres.

$$\lambda = \frac{300,000}{f}$$

when f = kcs., λ = metres.

$$\lambda = \frac{300}{f}$$

when f = mcs., λ = metres.

Dynamic Resistance at Resonance

$$R = 2\pi f L Q \text{ or } R = \frac{L}{C r}$$

when L = henrys, C = farads and r = HF resistance at frequency concerned.

Magnification of a Tuned Circuit

$$Q = \frac{2\pi f L}{r}$$

when L = henrys, C = farads and r = HF resistance at frequency concerned.

Selectivity of a Tuned Circuit

$$\text{Voltage on tuned circuit} = \frac{100}{\sqrt{1 + 4Q^2 \left(\frac{\delta f}{f}\right)}} \% \text{ of voltage at resonance}$$

when f = frequency of resonance in cycles,
 δf = amount of detuning in cycles, Q = magnification of tuned circuit.

Optimum Coupling

(for max. signal)

$$M = \frac{\sqrt{r_1 r_2}}{2\pi f}$$

when $r_1 r_2$ = HF resistance at resonance and f = frequency in cps. of resonance.

Coefficient of Coupling

$$k = \frac{M}{\sqrt{L_1 L_2}}$$

when M = mutual inductance, L_1, L_2 = inductance in henrys.

H.F. Transformer Ratio

$$n^2 = \frac{R}{R_a}$$

when n = ratio, R_a = valve impedance and R = dynamic resistance.

LC Constant

$$LC = \frac{25330}{f^2}$$

when L = μ H, C = $\mu\mu$ F and f = mcs.

Bandpass Tuning, Peak Separation

Inductive Coupling

$$\text{Peak separation} = \frac{\sqrt{(2\pi f)^2 M^2 -}}{2\pi L}$$

when M = mutual inductance in henrys, r = HF resistance in ohms, L = inductance in henrys and C_m = coupling capacity in farads.

Capacity Coupling

$$\text{Peak separation} = \frac{\sqrt{\frac{1}{(2\pi f)^2 C_m^2} - r^2}}{2\pi L}$$

Time Constants

RC = time in seconds for voltage across a discharging condenser to fall to 37% approx. of its original value

$\frac{L}{R}$ = time in seconds for current to reach 63% approx. of its ultimate steady value

when R = ohms, L = henrys and C = farads.

RC = time in seconds for a charging condenser to reach 63% approx. of its ultimate voltage

$\frac{L}{R}$ = time in seconds for a current to fall from its maximum steady value to 37% approx. of that value

H.F. Feeders

$$Z_o = \sqrt{\frac{L}{C}}$$

when Z_o = characteristic impedance. L = henrys and C = farads per unit length.

$$Z_o \text{ of twin feeder} = \frac{276, \log_{10} \frac{2D}{d}}{\sqrt{K}}$$

when d = diam. of wire, D = mean distance between conductors, K = dielectric constant.

$$Z_o \text{ of concentric feeder} = \frac{138, \log_{10} \frac{D}{d}}{\sqrt{K}}$$

when D = internal diam. of outer conductor, d = diam. of inner conductor and K = dielectric constant.

Aerials

$$\text{Length of } \frac{1}{2} \lambda \text{ aerial in feet} = \frac{468}{f(\text{mcs.})} \quad \text{Length of } \frac{1}{4} \lambda \text{ aerial in feet} = \frac{234}{f(\text{mcs.})} \quad \text{Length of long wire aerial in feet} = \frac{492(N - .05)}{f(\text{mcs.})}$$

when N = number of $\frac{1}{2}$ waves on aerial.

Valve Characteristics

$$\text{Impedance} = \frac{\text{Change in anode volts}}{\text{Change in anode current}}$$

$$\text{Amp. factor} = \frac{\text{Change in anode volts}}{\text{Change in grid volts}}$$

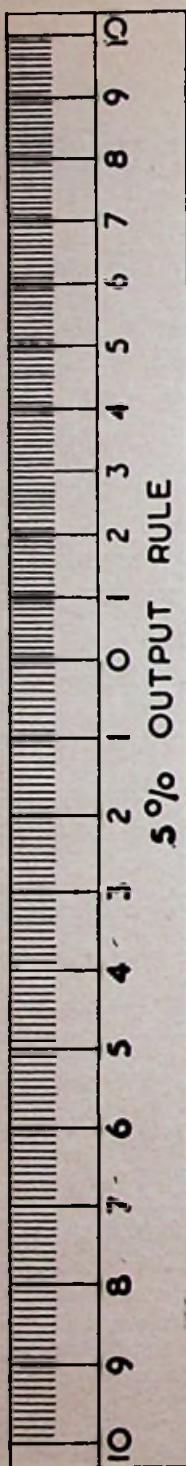
Relationship between impedance (R_a), amplification factor (μ) and slope (g_m) is :

$$g_m = \frac{\mu}{R_a}, \quad \mu = g_m R_a \quad \text{and} \quad R_a = \frac{\mu}{g_m}$$

when units are volts, amps. and ohms and g_m is in amps. per volt.

$$\text{Slope} = \frac{\text{Change in anode current}}{\text{Change in grid volts}}$$

$$\text{Slope} = \frac{\text{Amp. factor}}{\text{Impedance}}$$



Power Output and Distortion of Triode Valves

5% second harmonic distortion is present when the distance $I_{\max} - I_o$ is $\frac{1}{5}$ of the distance $I_o - I_{\min}$ (I_o is the operating or bias point). The use of a special ruler to satisfy this condition is explained on page 193, Vol. I, of *Modern Practical Radio and Television*. Suitable dimensions for making such a ruler are given on the left; any transparent material may be used, and great accuracy must be observed.

$$\text{Output watts} = \frac{1}{8} (I_{\max} - I_{\min})(V_{\max} - V_{\min})$$

$$\text{Load resistance} = \frac{V_{\max} - V_{\min}}{I_{\max} - I_{\min}}$$

when V = volts and I = current in amps.

Power Output, Optimum Load and Bias by Formulae

$$\text{Watts output} = 0.41 \mu k \left(\frac{V_a}{\mu} \right)^{\frac{2}{3}} \text{ approx.} \quad \text{Optimum load} = 1.9 \frac{\mu}{k} \left(\frac{V_a}{\mu} \right)^{-\frac{1}{3}}$$

when V_a = anode voltage in volts, μ = amplification factor and k = as below.

$$k = \frac{I}{\left(\frac{V_a - V_g}{\mu} \right)^{\frac{2}{3}}}$$

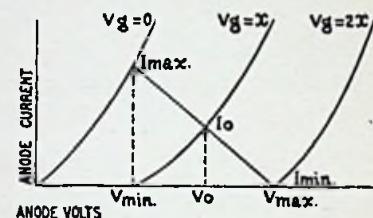
when V_a = anode voltage in volts, V_g = grid voltage in volts, I_a = anode current in amps.

Voltage Amplification (Stage Gain)

$$\text{Stage gain} = \frac{\mu Z}{Z + R_a}$$

when Z = impedance of anode load, μ = amplification factor and R_a = valve impedance. (If Z is inductive or capacitive, Z and R_a must be added vectorially.)

Diagrammatic representation of anode-volt/anode-current curve. I_o is the operating point.



$$\% \text{ 2nd harmonic distortion} = \frac{\frac{I_{\max} + I_{\min}}{2} - I_o}{I_{\max} - I_{\min}} \times 100\%$$

Units and value of the constant k are the same as those given, left.

$$\text{Approx. grid bias} = \frac{V_g \times 1,000}{(I_a + I_{sg})n}$$

when V_g = grid bias in volts, I_a = anode current in mA, I_{sg} = screen current in mA and n = number of valves using common bias resistor.

Output Transformer Ratio

$$N = \sqrt{\frac{R_o}{Z}}$$

When N = transformer turns ratio, R_o = optimum load of valve and Z = loudspeaker impedance.

R.C. Coupled A.F. Amplifier Design

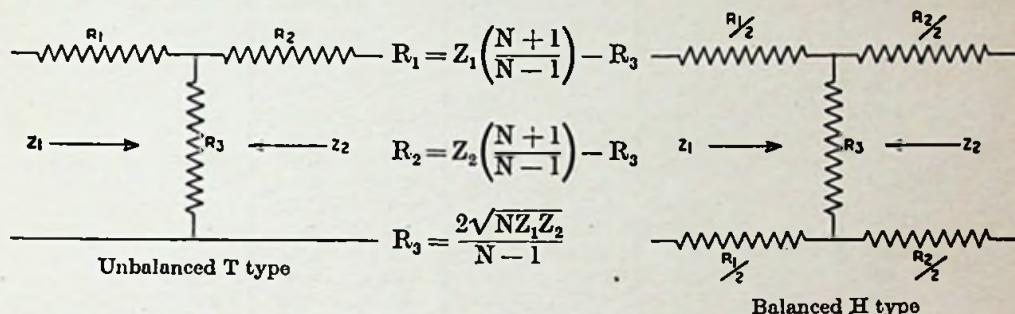
$$\text{Stage gain at medium frequencies} = \frac{\mu R}{R + R_a} = A_{mf}$$

$$\text{Stage gain at high frequencies} = \frac{A_{mf}}{\sqrt{1 + (2\pi f)^2 C_1^2 r^2}}$$

$$\text{when } R = \frac{R_1 R_2}{R_1 + R_2} \quad r = \frac{R R_a}{R + R_a} \quad \rho = R_2 + \frac{R_1 R_a}{R_1 + R_a}$$

and when R_a = valve impedance, R_1 = anode load resistance in ohms, R_2 = grid leak resistance in ohms, C_1 = total shunt capacity in farads, C_2 = coupling capacity in farads, μ = valve amplification factor, A_{mf} = see top left, ρ = see bottom right.

Attenuation Network Design



when Z = impedance in ohms, R = resistance in ohms and N = the ratio of power absorbed (by the attenuator) from the source of the power to the output power delivered,

Turns Required for a Given Inductance and Size of Coil Former

$$n = \sqrt{\left(\frac{L}{a}\right)\left(\frac{b}{a}\right)39.88\left(\frac{1}{K}\right)} \quad \text{when } n = \text{required number of turns, } L = \mu\text{H, } a = \text{diameter of coil in inches, } b = \text{length in inches and } K = \text{correction, for which see below.}$$

Inductance of a Single Layer Coil

$$L = \frac{0.0395a^2n^2}{b} K \quad \text{when } L = \mu\text{H, } a = \text{radius of coil measured from the axis to centre of wire, } n = \text{number of turns, } b = \text{length, } K = \text{correction as below (measurements in inches).}$$

$$\text{When } \frac{D}{B} = .1 \quad .25 \quad .5 \quad .75 \quad 1 \quad 1.5 \quad 2 \quad 2.5 \quad 3 \quad 3.5 \quad 4 \quad 4.5$$

$$K = .96 \quad .9 \quad .82 \quad .75 \quad .69 \quad .59 \quad .526 \quad .472 \quad .429 \quad .394 \quad .365 \quad .34$$

when D = mean diameter and B = length of winding.

Capacity of a Fixed Condenser

$$C = \frac{0.0885AK(n-1)}{d} \quad \text{when } A = \text{overlapping area of metal plates in square centimetres, } K = \text{dielectric constant, } n = \text{number of metal plates. } d = \text{distance between plates in centimetres.}$$

GREEK LETTERS.—Greek letters used in these pages of formulæ are pronounced as follows : ρ = rho, ϕ = phi, π = pi, λ = lambda, μ = mu and δ = delta.

**READY-WORKED FREQUENCY—WAVELENGTH
CONVERSION TABLE**

<i>f</i> in Mcs.	Metres	<i>f</i> in Mcs.	Metres	<i>f</i> in Kcs.	Metres	<i>f</i> in Kcs.	Metres	<i>f</i> in Kcs.	Metres
70	4.29	14	21.43	3,000	100	830	361.4	505	594.1
69	4.35	13.5	22.22	2,950	101.7	820	365.8	500	600
68	4.41	13	23.08	2,900	103.5	810	370.7	495	606
67	4.48	12.5	24	2,850	105.3	800	375	490	612.3
66	4.55	12	25	2,800	107.1	790	379.7	485	618.6
65	4.62	11.5	26.08	2,750	109.1	780	384.9	480	625
64	4.69	11	27.27	2,700	111.1	770	389.6	475	631.6
63	4.76	10.5	28.57	2,650	113.2	760	394.8	470	638.3
62	4.84	10	30	2,600	115.4	750	400	465	645.2
61	4.92	9.5	31.58	2,550	117.7	745	402.9	460	652.2
60	5	9	33.33	2,500	120	740	405.4	455	659.3
59	5.09	8.5	35.29	2,450	122.5	735	408.2	450	666.7
58	5.17	8	37.5	2,400	125	730	411	440	681.8
57	5.26	7.75	38.71	2,350	127.7	725	413.8	430	697.7
56	5.36	7.5	40	2,300	130.4	720	416.7	420	714.3
55	5.45	7.25	41.38	2,250	133.3	715	419.6	410	731.8
54	5.56	7	42.86	2,200	136.4	710	422	400	750
53	5.66	6.75	44.46	2,150	139.5	705	425.5	390	769.8
52	5.71	6.5	46.15	2,100	142.9	700	428.6	380	789.5
51	5.88	6.25	48	2,050	146.4	695	431.7	370	810.8
50	6	6	50	2,000	150	690	434.8	360	833.4
49	6.12	5.9	50.85	1,950	153.9	685	438	350	857.2
48	6.25	5.8	51.73	1,900	157.9	680	441.2	340	882.5
47	6.38	5.7	52.63	1,850	162.2	675	444.5	330	909.1
46	6.52	5.6	53.57	1,800	166.7	670	448.8	320	937
45	6.67	5.5	54.55	1,750	171.4	665	451.1	310	967.7
44	6.82	5.4	55.55	1,700	176.5	660	454.6	300	1,000
43	6.98	5.3	56	1,650	181.8	655	458	295	1,017
42	7.14	5.2	57.69	1,600	187.5	650	461.5	290	1,034.5
41	7.32	5.1	58.83	1,550	193.6	645	465.1	285	1,052.6
40	7.5	5	60	1,500	200	640	468.5	280	1,071.4
39	7.7	4.9	61.23	1,450	206.0	635	472.4	275	1,090.9
38	7.9	4.8	62.5	1,400	214.3	630	476.2	270	1,111.1
37	8.11	4.7	63.83	1,350	222.2	625	480	265	1,132
36	8.33	4.6	65.22	1,300	230.8	620	483.9	260	1,153.8
35	8.57	4.5	66.66	1,250	240	615	487.8	255	1,176.5
34	8.82	4.4	68.18	1,200	250	610	491.8	250	1,200
33	9.09	4.3	69.77	1,150	260.9	605	495.9	245	1,224.5
32	9.37	4.2	71.43	1,100	272.7	600	500	240	1,250
31	9.68	4.1	73.18	1,050	285.7	595	504.2	235	1,276.6
30	10	4	75	1,000	300	590	508.5	230	1,304.3
29	10.35	3.9	76.98	990	303	585	512.8	225	1,333.3
28	10.71	3.8	78.95	980	306.2	580	517.2	220	1,363.6
27	11.11	3.75	80	970	309.3	575	521.7	215	1,395.4
26	11.54	3.7	81.08	960	312.5	570	526.3	210	1,428.5
25	12	3.65	82.19	950	315.8	565	531	205	1,463.5
24	12.5	3.6	83.34	940	319.2	560	535.7	200	1,500
23	13.04	3.55	84.39	930	322.9	555	540.5	195	1,538.5
22	13.04	3.5	85.72	920	326.1	550	545.5	190	1,570
21	14.29	3.45	86.90	910	329.7	545	550.6	185	1,621.6
20	15	3.4	88.25	900	333.3	540	555.0	180	1,666.7
19	15.79	3.35	89.56	890	337.1	535	560.8	175	1,714.3
18	16.67	3.3	90.91	880	340.9	530	566	170	1,764.7
17	17.65	3.25	92.31	870	344.9	525	571.4	165	1,818.2
16	18.75	3.2	93.7	860	348.9	520	576.0	160	1,875
15	20	3.1	96.77	850	352.9	515	582.5	155	1,935.5
14.5	20.69	3	100	840	357.1	510	588.3	150	2,000

READY-WORKED COIL WINDING TABLES

MEDIUM WAVES

Former Diam. (inches)	175 Microhenrys *				200 Microhenrys †				225 Microhenrys ‡			
	S.W.G.	Type	Turns	Length (inches)	S.W.G.	Type	Turns	Length (inches)	S.W.G.	Type	Turns	Length (inches)
1.25	32	S.S.C.	85	1.05	32	S.S.C.	94	1.15	32	S.S.C.	104	1.26
1.25	30	D.S.C.	93	1.38	30	D.S.C.	103	1.53	30	D.S.C.	112	1.72
1.5	30	D.S.C.	74	1.1	30	D.S.C.	82	1.22	30	D.S.C.	90	1.35
1.75	30	D.S.C.	63	.95	30	D.S.C.	68	1	30	D.S.C.	76	1.15
1.75	28	D.S.C.	67	1.16	28	D.S.C.	73	1.25	28	D.S.C.	80	1.4
2	30	D.S.C.	54	.8	30	D.S.C.	59	.89	30	D.S.C.	65	.97
2	28	D.S.C.	58	1	28	D.S.C.	64	1.15	28	D.S.C.	70	1.23
2.25	28	D.S.C.	52	.93	28	D.S.C.	57	1	28	D.S.C.	62	1.1
2.25	26	D.C.C.	70	1.95	26	D.C.C.	77	2.1	26	D.C.C.	87	2.39
2.5	28	D.S.C.	47	.84	28	D.S.C.	51	.9	28	D.S.C.	56	1
2.5	24	D.C.C.	58	1.85	24	D.C.C.	65	2.05	24	D.C.C.	72	2.25

Note.—* For normal broadcast coverage where a trimmer is employed, use 175 μH with .0005 mfd.

† Use with .0005 mfd. for normal broadcast coverage.

‡ Use with .0004 mfd. for normal broadcast coverage.

LONG WAVES

Former Diam. (inches)	2,100 Microhenrys *				3,000 Microhenrys †			
	S.W.G.	Type	Grooves‡	Turns per groove	S.W.G.	Type	Grooves‡	Turns per groove
1	36	Enamel	4	92	36	Enamel	5	95
1	36	Enamel	5	80	36	Enamel	6	85
1.5	36	Enamel	3	80	36	Enamel	3	96
1.5	34	Enamel	5	60	36	Enamel	4	78
2	34	D.S.C.	4	51	36	D.S.C.	4	60
2	34	D.S.C.	5	45	36	D.S.C.	5	53

Note.—* For normal long-wave coverage use .0005 mfd.

† For normal long-wave coverage use .0004 mfd.

‡ Grooves must be $\frac{1}{8}$ -inch deep, $\frac{1}{16}$ -inch wide and be $\frac{1}{2}$ -inch apart; such formers may be purchased ready-made in ebonite or other suitable insulating material.

READY-WORKED L × C CONSTANTS FOR RADIO AND AUDIO FREQUENCIES

Radio Frequencies L in μH and C in μF		Radio Frequencies—continued		Audio Frequencies L in H and C in μF	
F in Kcs.	L × C	F in Kcs.	L × C	F in cps.	L × C
30,000	.0000282	680	.05477	25	40.545
15,000	.0001129	660	.05815	50	10.136
10,000	.000253	640	.06184	100	2.533
7,300	.0004753	630	.06382	200	.633
6,000	.000704	620	.06484	400	.158
5,000	.001014	610	.06807	600	.07
1,000	.02533	600	.07036	800	.04
950	.02800	590	.07276	1,000	.025
900	.03127	580	.07529	2,000	.0063
875	.03308	570	.07790	3,000	.0028
850	.03506	560	.08076	4,000	.0016
825	.03721	550	.08373	4,500	.00125
800	.03957	465	.11715	5,000	.001
780	.04163	300	.28145	6,000	.0007
760	.04385	260	.40545	7,000	.0005
740	.04625	200	.63325	8,000	.0004
720	.04891	170	.87046	9,000	.00031
700	.05140	150	1.1258	10,000	.00025

Note.—Many formulae incorporate $\sqrt{L \times C}$; above figures are L × C, the square-root of which can be taken when necessary.

COPPER WIRE TABLE

Imperial or British Standard Wire Gauge	U.S.A. Wire Gauge (B. & S.)	Diameter in inches	Diameter in millimetres	*Ohms per thousand feet	Feet per lb.	† Turns per inch						
						S.W.G.	A.W.G. (B. & S.)	Bare	Rharn.	S.C.C.	D.C.C.	S.S.C.
10	-	.128	3.25	.0219	20.16	7.81	7.63	7.35	7.04	-	-	-
11	-	.116	2.946	.0219	24.55	8.02	8.33	8.07	7.69	-	-	-
12	0	.114	2.906	.0219	26.23	8.74	8.58	8.23	7.91	-	-	-
13	10	.104	2.64	.042	30.54	0.02	0.26	8.93	8.48	-	-	-
13	11	.1019	2.558	.0089	31.82	0.81	0.61	9.26	8.85	-	-	-
13	12	.092	2.337	.204	30.01	10.87	10.42	10.00	9.43	-	-	-
13	13	.092	2.304	.120	40.12	11.02	10.7	10.4	9.98	-	-	-
13	14	.09074	2.053	.1588	50.59	12.37	12.00	11.6	11.07	-	-	-
13	15	.08081	2.053	.1588	61.80	12.50	11.90	11.36	10.64	-	-	-
14	16	.0720	1.828	.0219	2.000	63.80	13.89	13.33	12.50	-	-	-
14	16	.064	1.63	.0219	2.488	80.05	16.59	14.81	14.08	13.16	14.93	14.71
14	17	.064	1.628	.0219	2.625	80.44	16.61	15.1	14.4	13.59	-	-
15	17	.05707	1.450	.184	101.4	17.52	16.9	16.1	15.00	-	-	-
15	17	.056	1.422	.184	105.4	17.86	16.95	16.87	14.71	16.95	16.07	16.07
16	18	.05082	1.291	.0219	4.016	127.9	19.68	18.9	16.5	18.9	18.2	18.2
16	18	.048	1.22	.0219	4.422	143.3	20.83	19.72	18.18	16.95	20.0	19.01
17	18	.04626	1.160	.0219	5.064	161.3	22.1	21.2	19.8	18.2	21.1	20.2
18	18	.04030	1.024	.0219	6.385	203.4	24.8	23.7	22.0	20.0	23.6	22.6
18	19	.040	1.016	.0219	6.368	206.4	25.0	24.7	21.28	19.61	23.81	23.26
19	20	.036	.914	.0219	7.860	254.8	27.78	26.97	23.81	21.28	26.32	25.04
19	21	.03589	.9116	.0219	8.051	256.5	27.8	26.5	24.4	22.0	26.3	25.0
20	21	.032	.8138	.0219	9.950	322.6	31.25	28.15	26.32	23.26	28.67	28.67
21	21	.03196	.8118	.0219	10.15	323.4	31.3	29.5	27.0	24.1	29.4	27.7
21	22	.02846	.7230	.0219	12.80	407.8	35.1	33.1	29.8	26.3	32.7	30.7
22	22	.028	.711	.0219	12.897	421.2	35.71	33.33	29.41	25.64	33.33	32.26
22	23	.02535	.6438	.0219	16.14	514.2	39.4	37.0	33.5	29.5	36.6	34.1
23	23	.024	.6096	.0219	17.69	573.4	41.67	38.91	34.48	29.41	38.46	37.04
23	24	.02257	.5733	.0219	20.30	648.4	44.3	41.4	36.9	32.1	40.6	37.5
24	24	.022	.5569	.0219	682.6	45.45	42.37	37.04	31.25	42.55	40.00	42.55
24	25	.0201	.5106	.0219	25.67	817.7	49.7	46.5	40.6	34.9	45.2	41.4
25	25	.020	.508	.0219	25.47	823.8	60.00	46.51	40.00	33.33	43.48	43.48
26	26	.018	.457	.0219	31.45	1.019	55.56	51.55	43.48	35.71	51.81	48.78
26	26	.0179	.4547	.0219	32.37	1.031	56.8	52.0	44.6	37.8	50.0	45.6
26	27	.0164	.4105	.0219	1.229	37.88	1.080	1.080	46.73	37.88	52.91	52.91
26	27	.01594	.4049	.0219	1.300	40.81	62.7	58.4	49.0	40.9	55.8	50.0
27	28	.0148	.376	.0219	40.52	1.608	67.67	62.60	50.51	40.32	62.11	57.80
27	28	.0142	.3606	.0219	61.47	1.639	70.4	65.3	53.4	44.0	61.7	54.9
27	29	.0136	.3454	.0219	55.09	1.786	73.53	67.67	53.76	42.37	62.11	52.11

* Resistance of U.S.A. gauges measured at 68° F. and S.W.G. gauges at 60° F.

† Apparent inconsistencies in turns per inch between similar British and U.S.A. gauges are due to differences in thickness of insulation.

COPPER WIRE TABLE (contd.)

R.C.C. Imperial or British Standard Wire Gauge in	U.S.A. Wire Gauge	Diameter in inches	Diameter in millimetres	*Ohms per thousand feet	Feet per lb.	† Turns per inch					
						—	—	—	—	—	—
S.W.G.	A.W.G. (B. & S.)	—	—	—	—	Bare	Enam.	S.C.C.	D.C.C.	S.S.C.	D.S.C.
28	.01264	.3211	0.490	2.067	82.8	73.5	58.4	47.3	68.4	60.2	
30	—	.0124	0.316	0.627	2.148	80.65	74.03	67.47	44.64	72.90	67.11
31	—	.0116	0.2846	0.757	2.465	86.21	79.37	60.24	40.30	77.52	70.92
32	—	.01126	0.2859	0.8183	2.007	88.8	81.9	63.2	50.5	75.1	65.3
33	—	.0108	0.274	0.874	2.822	92.59	85.47	63.29	48.0	82.64	76.19
34	—	.01003	.02540	101.4	3.302	100.0	91.74	66.67	50.0	88.60	80.0
35	—	.0092	.0234	103.2	3.287	99.7	92.5	68.9	54.0	83.3	71.4
36	—	.00883	.02208	120.4	3.901	108.0	99.0	70.42	52.08	95.24	86.47
37	—	.0082	.02082	130.1	4.445	112.0	103.0	74.4	57.4	01.7	77.5
38	—	.0079	.02050	144.4	4.082	119.0	110.0	80.65	57.47	103.0	91.74
39	—	.0072	.0193	164.1	5.227	125.8	114.0	80.80	60.0	100.0	83.3
40	—	.0071	.01798	200.9	6.718	131.6	118.0	88.21	60.25	112.6	99.0
41	—	.0068	.01727	220.4	7.143	141.2	129.0	86.2	64.1	108.0	90.0
42	—	.0063	.01601	260.9	8.310	147.0	135.0	99.21	63.29	123.6	107.6
43	—	.006	.0162	263.0	0.74	166.7	144.0	92.5	67.5	120.0	97.0
44	—	.0056	.01426	329.0	10.480	178.0	166.5	100.0	66.67	137.0	118.0
45	—	.0052	.01320	370.8	12.210	102.3	175.4	98.9	70.0	131.0	104.0
46	—	.0050	.01270	414.8	13.210	200.0	181.0	108.7	70.42	153.8	130.0
47	—	.0048	.0122	442.2	14.330	208.0	185.7	111.0	76.9	142.0	111.0
48	—	.0044	.01117	520.3	17.080	227.0	206.0	113.0	72.46	164.0	137.0
49	—	.00445	.01131	532.1	16.060	224.0	204.0	206.0	—	178.6	161.6
50	—	.004	.0102	636.8	20.640	250.0	222.0	170.0	80.0	153.0	117.0
51	—	.00396	.0097	659.6	21.010	252.0	227.0	125.0	—	192.3	161.2
52	—	.0036	.00914	786.3	25.460	277.8	255.0	125.0	83.3	166.0	126.0
53	—	.00353	.00897	831.8	26.500	283.0	268.0	133.0	—	208.3	172.3
54	—	.0032	.0081	905.0	32.260	312.6	277.7	113.0	86.9	181.0	133.0
55	—	.00314	.00799	1,048.0	33.410	318.0	285.0	140.0	—	227.3	185.2
56	—	.0028	.00711	1,299.7	42.120	357.1	315.1	140.0	90.0	198.0	140.0
57	—	.00275	.00698	1,370.0	43.700	303.0	327.0	125.0	—	250.0	200.0
58	—	.0025	.00635	1,680.0	52.800	400.0	378.0	125.0	—	—	—
59	—	.0024	.0061	1,789.0	57.340	416.7	388.3	125.0	—	277.7	217.4
60	—	.00235	.00571	2,050.0	65.300	444.0	421.0	125.0	—	—	—
61	—	.002	.00504	2,547.0	82.680	500.0	435.5	125.0	312.5	238.0	238.0
62	—	.00175	.00404	2,600.0	82.600	471.0	453.0	125.0	—	—	—

* Resistance of U.S.A. gauges measured at 68° F. and S.W.G. gauges at 60° F.

† Apparent inconsistencies in turns per inch between similar British and U.S.A. gauges are due to differences in thickness of insulation.

MISCELLANEOUS WIRE TABLES

FUSE-WIRE TABLE

Fusing Current in amps.	Tin		Tin Alloy (75% Lead, 25% Tin)		Tinned Copper		Lead	
	Diameter in inches	S.W.G.	Diameter in inches	S.W.G.	Diameter in inches	S.W.G.	Diameter in inches	S.W.G.
1	.0072	37	.0084	35	—	—	.0084	35
2	.0113	31	.0136	29	—	—	.0124	30
3	.0148	28	.0164	27	—	—	.0164	27
4	.018	26	.02	25	—	—	.02	25
5	.02	25	.024	23	—	—	.024	23
6	.024	23	.032	21	—	—	—	—
10	.032	21	.048	18	.01	33	.036	20
15	.044	—	.064	16	.0136	29	.048	18
20	.056	17	—	—	.0146	27	.056	17
25	.064	16	—	—	—	—	.072	15
30	.072	15	—	—	.022	24	.08	14

The above figures are approximate owing to the desirability of using S.W.G. and round figures in amps.; they are, however, sufficiently accurate for practical purposes. Sixty per cent. may be taken as the safe normal current.

FLEX

Description	Sectional Area in sq. ins. (nominal)	Maximum Current in amps.	Description	Sectional Area in sq. ins. (nominal)	Maximum Current in amps.
14/-0076	.0006	2	70/-0076	.003	10
23/-0076	.001	3	110/-0076	.0048	15
40/-0076	.0017	5	102/-0076	.007	20

RESISTANCE WIRE

S.W.G.	Diameter (inches)	Eureka* (ohms per yard)	Nichrome † (ohms per yard)	S.W.G.	Diameter (inches)	Eureka * (ohms per yard)	Nichrome † (ohms per yard)
10	.128	.053	.125	30	.0124	5.633	12.982
12	.104	.08	.189	31	.0116	6.436	14.825
13	.092	.1023	.242	32	.0108	7.427	17.102
14	.08	.1353	.32	33	.01	8.662	19.960
15	.072	.1671	.395	34	.0092	10.23	23.57
16	.064	.2115	.486	35	.0084	12.27	28.277
17	.056	.2762	.636	36	.0076	15	34.563
18	.048	.376	.865	37	.0068	18.74	43.166
19	.04	.541	1.246	38	.006	24.05	55.374
20	.036	.668	1.539	39	.0052	32.03	73.722
21	.032	.846	1.957	40	.0048	37.6	86.486
22	.028	1.105	2.546	41	.0044	44.74	100.15
23	.024	1.504	3.463	42	.0040	54.13	124.7
24	.022	1.789	4.118	43	.0036	66.68	153.875
25	.02	2.166	4.979	44	.0032	84.61	194.94
26	.018	2.673	6.156	45	.0028	110.5	254.6
27	.0164	3.221	7.422	46	.0024	150.4	346.3
28	.0148	4.119	9.107	47	.002	216.6	497.89
29	.0136	4.684	10.793	48	.0016	338.4	779.52

* Values are approximate and are based on an alloy of 40% nickel and 60% copper. This wire should not be used above 575° F. (300° C.). It has an extremely low temperature coefficient.

† The resistance of nichrome varies considerably; figures given are the average of several different manufacturers, all being alloys of approximately 80% nickel and 20% chromium. Alloys containing a small percentage of iron have a slightly lower resistance; temperature coefficient is low, averaging + 0.0001 per C°.

READY-WORKED VOLTAGE, CURRENT AND RESISTANCE TABLE
VOLTAGE DROP IN VOLTS

	1	2	4	6	8	10	14	18	22	26	30	34	38	42	46	50	60	70	80	100
1	1,000	2,000	4,000	6,000	8,000	10,000	14,000	18,000	22,000	26,000	30,000	34,000	38,000	42,000	46,000	50,000	60,000	70,000	80,000	100,000
2	600	1,000	2,000	3,000	4,000	5,000	7,000	9,000	11,000	13,000	15,000	17,000	19,000	21,000	23,000	25,000	30,000	35,000	40,000	50,000
3	333	667	1,332	2,000	2,664	3,333	4,007	6,000	7,333	8,067	10,000	11,333	12,067	14,000	15,333	16,067	20,000	23,333	26,667	33,333
4	250	500	1,000	1,500	2,000	2,500	3,500	4,500	5,500	6,500	7,500	8,500	9,500	10,500	11,500	12,500	15,000	17,500	20,000	25,000
5	200	400	800	1,200	1,600	2,000	2,800	3,600	4,400	5,200	6,000	6,800	7,200	8,400	9,200	10,000	12,000	14,000	16,000	20,000
6	167	333	667	1,000	1,333	1,667	2,007	2,333	3,000	3,807	4,333	5,000	5,667	6,333	7,000	7,667	8,333	10,000	11,667	13,333
7	143	283	572	857	1,143	1,430	2,000	2,571	3,142	3,714	4,286	4,857	5,429	6,000	6,571	7,143	8,571	10,000	11,429	14,286
8	125	250	500	750	1,000	1,250	1,750	2,250	2,750	3,250	3,750	4,250	4,750	5,250	5,750	6,250	7,500	8,750	10,000	12,500
9	111	222	444	667	889	1,111	1,556	2,000	2,444	2,889	3,333	3,778	4,222	4,667	5,111	6,556	6,667	7,778	8,889	11,111
10	100	200	400	600	800	1,000	1,400	1,800	2,200	2,600	3,000	3,400	3,800	4,200	4,600	5,000	6,000	7,000	8,000	10,000
11	83	167	333	500	667	833	1,167	1,500	1,833	2,167	2,500	2,833	3,107	3,500	3,833	4,167	5,000	5,833	6,007	8,333
12	71	143	283	428	572	714	1,000	1,286	1,571	1,857	2,143	2,429	2,714	3,000	3,286	3,571	4,286	5,000	5,714	7,143
13	63	125	250	375	500	625	875	1,125	1,375	1,625	1,875	2,126	2,375	2,625	2,875	3,125	3,750	4,375	5,000	6,250
14	56	111	222	333	444	556	778	1,000	1,222	1,444	1,667	1,888	2,111	2,333	2,556	2,777	3,333	3,889	4,444	5,556
15	50	100	200	300	400	500	700	900	1,100	1,300	1,500	1,700	2,100	2,300	2,500	2,700	3,000	3,500	4,000	5,000
16	45	91	182	273	364	456	636	818	1,000	1,182	1,304	1,504	1,727	1,909	2,091	2,273	2,727	3,182	3,636	4,546
17	42	83	167	250	333	417	583	750	917	1,083	1,250	1,417	1,583	1,750	1,917	2,083	2,500	2,917	3,333	4,167
18	39	77	153	231	308	385	538	692	846	1,000	1,154	1,308	1,462	1,615	1,769	1,923	2,308	2,692	3,077	3,846
19	36	71	143	214	286	367	500	643	786	929	1,071	1,214	1,357	1,500	1,643	1,780	2,143	2,500	2,857	3,671
20	33	67	133	200	267	333	467	600	733	867	1,000	1,133	1,267	1,400	1,533	1,679	2,000	2,333	2,667	3,333
21	31	63	125	188	250	313	437	561	688	813	938	1,063	1,188	1,313	1,438	1,563	1,876	2,188	2,550	3,125
22	29	59	118	170	235	294	412	520	647	765	882	1,000	1,118	1,235	1,353	1,471	1,765	2,059	2,365	2,941
23	28	56	111	167	222	278	389	500	611	722	833	944	1,056	1,167	1,278	1,389	1,607	1,944	2,222	2,778
24	25	50	100	150	200	350	450	560	650	750	850	950	1,050	1,160	1,250	1,500	1,750	2,000	2,500	3,281
25	24	48	95	143	190	238	333	429	524	643	714	810	905	1,000	1,096	1,190	1,439	1,667	1,905	2,222
26	23	45	91	136	182	227	318	400	500	591	682	773	864	955	1,045	1,130	1,364	1,691	1,818	2,273
27	20	53	105	168	211	263	368	474	678	684	780	806	1,000	1,105	1,211	1,316	1,679	1,842	2,105	2,632
28	19	43	87	130	174	217	304	391	478	565	652	739	826	913	1,000	1,067	1,304	1,622	1,739	2,174
29	18	42	83	125	167	208	292	379	458	542	625	708	792	875	958	1,042	1,250	1,458	1,667	2,083
30	20	40	80	120	160	200	280	360	440	520	600	680	760	840	920	1,000	1,200	1,400	1,600	2,000
31	19	37	74	111	146	185	259	333	407	481	566	630	704	778	852	926	1,111	1,291	1,481	1,892
32	18	36	71	107	143	170	250	321	393	464	536	607	671	750	821	893	1,071	1,250	1,429	1,780
33	17	34	69	103	138	172	241	310	379	457	517	586	656	724	793	862	1,034	1,207	1,379	1,724
34	17	33	67	100	133	167	233	300	367	433	500	567	633	700	767	833	1,000	1,167	1,333	1,667

CURRENT IN AMPS

READY-WORKED RESISTANCE COLOUR CODE

<i>Ohms</i>	<i>Body</i>	<i>Tip</i>	<i>Dot</i>	<i>Ohms</i>	<i>Body</i>	<i>Tip</i>	<i>Dot</i>
10	Brown	Black	Black	820	Grey	Red	Brown
11	Brown	Brown	Black	910	White	Brown	Brown
12	Brown	Red	Black	1,000	Brown	Black	Red
13	Brown	Orange	Black	1,100	Brown	Brown	Red
15	Brown	Green	Black	1,200	Brown	Red	Red
16	Brown	Blue	Black	1,300	Brown	Orange	Red
18	Brown	Grey	Black	1,500	Brown	Green	Red
20	Red	Black	Black	1,600	Brown	Blue	Red
22	Red	Red	Black	1,800	Brown	Grey	Red
24	Red	Yellow	Black	2,000	Red	Black	Red
27	Red	Violet	Black	2,200	Red	Red	Red
30	Orange	Black	Black	2,400	Red	Yellow	Red
33	Orange	Orange	Black	2,500	Red	Green	Red
36	Orange	Blue	Black	2,700	Red	Violet	Red
39	Orange	White	Black	3,000	Orange	Black	Red
43	Yellow	Orange	Black	3,300	Orange	Orange	Red
47	Yellow	Violet	Black	3,500	Orange	Green	Red
50	Green	Black	Black	3,600	Orange	Blue	Red
51	Green	Brown	Black	3,900	Orange	White	Red
56	Green	Blue	Black	4,000	Yellow	Black	Red
62	Blue	Red	Black	4,300	Yellow	Orange	Red
68	Blue	Grey	Black	4,500	Yellow	Green	Red
75	Violet	Green	Black	4,700	Yellow	Violet	Red
82	Grey	Red	Black	5,000	Green	Black	Red
91	White	Brown	Black	5,100	Green	Brown	Red
100	Brown	Black	Brown	5,600	Green	Blue	Red
110	Brown	Brown	Brown	6,000	Blue	Black	Red
120	Brown	Red	Brown	6,200	Blue	Red	Red
130	Brown	Orange	Brown	6,800	Blue	Grey	Red
150	Brown	Green	Brown	7,000	Violet	Black	Red
160	Brown	Blue	Brown	7,500	Violet	Green	Red
180	Brown	Grey	Brown	8,000	Grey	Black	Red
200	Red	Black	Brown	8,200	Grey	Red	Red
220	Red	Red	Brown	9,000	White	Black	Red
240	Red	Yellow	Brown	9,100	White	Brown	Red
250	Red	Green	Brown	10,000	Brown	Black	Orange
270	Red	Violet	Brown	11,000	Brown	Brown	Orange
300	Orange	Black	Brown	12,000	Brown	Red	Orange
330	Orange	Orange	Brown	13,000	Brown	Orange	Orange
350	Orange	Green	Brown	15,000	Brown	Green	Orange
360	Orange	Blue	Brown	16,000	Brown	Blue	Orange
390	Orange	White	Brown	18,000	Brown	Grey	Orange
400	Yellow	Black	Brown	20,000	Red	Black	Orange
430	Yellow	Orange	Brown	22,000	Red	Red	Orange
450	Yellow	Green	Brown	24,000	Red	Yellow	Orange
470	Yellow	Violet	Brown	27,000	Red	Violet	Orange
500	Green	Black	Brown	30,000	Orange	Black	Orange
510	Green	Brown	Brown	33,000	Orange	Orange	Orange
560	Green	Blue	Brown	35,000	Orange	Green	Orange
620	Blue	Red	Brown	36,000	Orange	Blue	Orange
680	Blue	Grey	Brown	39,000	Orange	White	Orange
700	Violet	Black	Brown	40,000	Yellow	Black	Orange
750	Violet	Green	Brown	43,000	Yellow	Orange	Orange

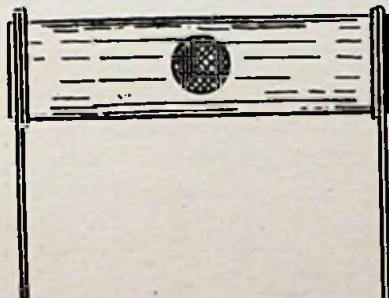
READY-WORKED RESISTANCE COLOUR CODE (*contd.*)

<i>Ohms</i>	<i>Body</i>	<i>Tip</i>	<i>Dot</i>	<i>Ohms</i>	<i>Body</i>	<i>Tip</i>	<i>Dot</i>
47,000	Yellow	Violet	Orange	390,000	Orange	White	Yellow
50,000	Green	Black	Orange	400,000	Yellow	Black	Yellow
51,000	Green	Brown	Orange	430,000	Yellow	Orange	Yellow
56,000	Green	Blue	Orange	450,000	Yellow	Green	Yellow
60,000	Blue	Black	Orange	470,000	Yellow	Violet	Yellow
62,000	Blue	Red	Orange	500,000	Green	Black	Yellow
68,000	Blue	Grey	Orange	510,000	Green	Brown	Yellow
70,000	Violet	Black	Orange	560,000	Green	Blue	Yellow
75,000	Violet	Green	Orange	600,000	Blue	Black	Yellow
82,000	Grey	Red	Orange	620,000	Blue	Red	Yellow
91,000	White	Brown	Orange	680,000	Blue	Grey	Yellow
100,000	Brown	Black	Yellow	750,000	Violet	Green	Yellow
110,000	Brown	Brown	Yellow	820,000	Grey	Red	Yellow
120,000	Brown	Red	Yellow	910,000	White	Brown	Yellow
130,000	Brown	Orange	Yellow	1 MΩ	Brown	Black	Green
150,000	Brown	Green	Yellow	1·1 MΩ	Brown	Brown	Green
160,000	Brown	Blue	Yellow	1·2 MΩ	Brown	Red	Green
180,000	Brown	Grey	Yellow	1·3 MΩ	Brown	Orange	Green
200,000	Red	Black	Yellow	1·5 MΩ	Brown	Green	Green
220,000	Red	Red	Yellow	1·6 MΩ	Brown	Blue	Green
230,000	Red	Orange	Yellow	1·8 MΩ	Brown	Grey	Green
240,000	Red	Yellow	Yellow	2 MΩ	Red	Black	Green
250,000	Red	Green	Yellow	2·2 MΩ	Red	Red	Green
270,000	Red	Violet	Yellow	2·4 MΩ	Red	Yellow	Green
280,000	Red	Grey	Yellow	2·7 MΩ	Red	Violet	Green
300,000	Orange	Black	Yellow	3 MΩ	Orange	Black	Green
330,000	Orange	Orange	Yellow	4 MΩ	Yellow	Black	Green
350,000	Orange	Green	Yellow	5 MΩ	Green	Black	Green
360,000	Orange	Blue	Yellow	10 MΩ	Brown	Black	Blue

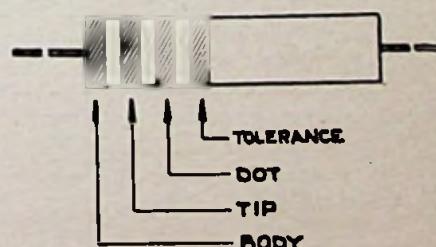
THE RESISTANCE COLOUR CODE (British and U.S.A.)

<i>Colour</i>	<i>Body</i>	<i>Tip</i>	<i>Dot</i>	<i>Colour</i>	<i>Body</i>	<i>Tip</i>	<i>Dot</i>
Black	0	0	·0	Green	5	5	00000
Brown	1	1	0	Blue	6	6	000000
Red	2	2	00	Violet	7	7	0000000
Orange	3	3	000	Grey	8	8	00000000
Yellow	4	4	0000	White	9	9	000000000

U.S.A.
only



The body colour denotes the first digit,
the colour of the tip the second digit,
and the colour of the spot the number
of noughts after the second digit.



The modified colour code used for resistances with axial leads.

READY-WORKED REACTANCE TABLE (in ohms)

<i>Inductance in Henrys</i>	30 cps.	50 cps.	100 cps.	400 cps.	1,000 cps.	3,000 cps.	5,000 cps.	
100	18,800	31,400	62,800	251,000	628,000	1,880,000	3,140,000	
75	14,130	23,550	47,100	188,100	471,000	1,413,000	2,355,000	
50	9,420	15,700	31,400	120,000	314,000	942,000	1,570,000	
25	4,710	7,850	15,700	62,800	157,000	471,000	785,000	
10	1,880	3,140	6,280	25,100	62,800	188,000	314,000	
1	188	314	628	2,510	6,280	18,800	31,400	
.1	18.8	31.4	6.28	.251	.628	1.880	3.140	
.01	1.88	3.14	.628	.251	.628	1.88	3.14	
.001	.188	.314	.0628	.0251	.0628	.188	.314	
								A.F. INDUCTIVE

<i>Capacity in mfd.</i>	30 cps.	50 cps.	100 cps.	400 cps.	1,000 cps.	3,000 cps.	5,000 cps.	
.00005					3.98 MΩ	3.18 MΩ	1.06 MΩ	637,000
.0001					1.59 MΩ	531,000	531,000	318,000
.00025					637,000	212,000	127,000	
.0003					531,000	174,000	106,000	
.0005	10.6 MΩ	6.37 MΩ	5.31 MΩ	3.18 MΩ	796,000	318,000	106,000	
.001	5.3 MΩ	3.18 MΩ	1.59 MΩ	1.59 MΩ	398,000	159,000	53,100	31,800
.005	1.06 MΩ	637,000	318,000	79,600	31,800	10,600	6,370	
.01	531,000	318,000	159,000	39,800	15,900	5,310	3,180	
.02	265,000	159,000	79,600	19,900	7,900	2,050	1,590	
.025	212,000	127,000	63,700	15,900	6,370	2,120	1,270	
.03	174,000	106,000	53,100	13,300	5,310	1,740	1,060	
.05	108,000	63,700	31,800	7,960	3,180	1,000	637	
.1	53,100	31,800	15,900	3,980	1,590	531	318	
.25	21,200	12,700	6,370	1,590	.637	212	127	
.5	10,600	6,370	3,180	.796	.318	106	63.7	
1.0	5,310	3,180	1,590	.398	.159	53.1	31.8	
2.0	2,650	1,590	.796	.199	.79.6	26.5	15.9	
25.0	212	127	63.7	15.9	6.37	2.12	1.27	
50.0	106	63.7	31.8	.796	.318	1.06	.637	
								A.F. CAPACITIVE

<i>Inductance</i>	100 Kcs.	250 Kcs.	465 Kcs.	500 Kcs.	1,000 Kcs.	1,500 Kcs.	5,000 Kcs.	
.1H	62,800	157,000	292,000	314,000	628,000	943,000	3,140,000	
.01H	6,280	15,700	29,200	31,400	62,800	94,300	314,000	
1,000 μH	628	1,570	2,920	3,140	6,280	9,430	31,400	
200 μH	126	315	484	630	1,260	1,890	6,300	
100 μH	62	157	292	314	628	943	3,140	
50 μH	31	78	146	157	314	476	1,570	
								R.F. INDUCTIVE

<i>Capacity in mfd.</i>	100 Kcs.	250 Kcs.	465 Kcs.	500 Kcs.	1,000 Kcs.	1,500 Kcs.	5,000 Kcs.	
.00001	159,000	636,000	34,200	31,800	15,900	10,600	3,180	
.00005	31,800	12,720	6,860	6,360	3,180	2,120	636	
.0001	15,900	6,360	3,420	3,180	1,590	1,060	318	
.0005	3,180	1,272	.685	.636	.318	.212	.636	
.001	1,590	.636	.342	.318	.159	.106	.318	
.005	.318	.127.2	.68.5	.63.6	.31.8	.21.2	.63.6	
.01	.159	.03.6	.34.2	.31.8	.15.9	.10.6	.31.8	
.05	.31.8	.12.72	.6.85	.6.36	.3.18	.2.12	.6.36	
.1	.15.9	.6.36	.3.42	.3.18	.1.59	.1.06	.318	
.25	.6.36	.2.544	.1.37	.1.272	.0.36	.424	.127	
.5	.3.18	.1.272	.6.85	.6.36	.3.18	.2.12	.664	
1.0	.1.59	.6.36	.3.42	.3.18	.1.59	.1.06	.32	
								R.F. CAPACITIVE

READY-WORKED DECIBEL TABLE

LOSS		db	GAIN	
Current or Voltage Ratio	Power Ratio		Current or Voltage Ratio	Power Ratio
1	1	0	1	1
.989	.977	.1	1.012	1.023
.977	.955	.2	1.023	1.047
.966	.933	.3	1.035	1.072
.955	.912	.4	1.047	1.096
.944	.891	.5	1.059	1.122
.933	.871	.6	1.072	1.148
.923	.851	.7	1.084	1.175
.912	.832	.8	1.096	1.202
.902	.813	.9	1.109	1.23
.891	.794	1	1.122	1.250
.871	.759	1.2	1.15	1.32
.851	.724	1.4	1.17	1.38
.832	.692	1.6	1.2	1.44
.813	.661	1.8	1.23	1.51
.794	.631	2	1.26	1.58
.750	.502	2.5	1.34	1.78
.708	.501	3	1.41	1.99
.668	.447	3.5	1.5	2.24
.631	.398	4	1.59	2.51
.596	.355	4.5	1.68	2.82
.562	.316	5	1.78	3.16
.531	.282	5.5	1.88	3.55
.501	.251	6	1.99	3.98
.473	.224	6.5	2.11	4.47
.447	.199	7	2.24	5.01
.422	.178	7.5	2.37	5.62
.398	.158	8	2.51	6.31
.376	.141	8.5	2.66	7.08
.355	.126	9	2.82	7.94
.335	.112	9.5	2.98	8.91
.316	.1	10	3.16	10
.282	.079	11	3.55	12.6
.251	.063	12	3.98	15.8
.224	.05	13	4.47	19.9
.199	.04	14	5.01	25.1
.158	.025	16	6.31	39.8
.126	.016	18	7.94	63.1
.1	.01	20	10	100
.056	.00316	25	17.78	316
.032	.001	30	31.62	1,000
.0178	.000316	35	56.23	3,160
.01	.0001	40	100	10,000
.0056	.0000316	45	177.8	31,600
.00316	.00001	50	316	100,000
.002	.00000316	55	562	316,000
.001	.000001	60	1,000	1,000,000
.0006	.000000316	65	1,770	3,160,000
.000316	.0000001	70	3,160	10,000,000
.0002	.0000000316	75	5,620	31,600,000
.0001	.00000001	80	10,000	100,000,000
.00006	.0000000032	85	17,800	316,000,000
.0000316	.000000001	90	31,600	1,000,000,000
.00002	.0000000003	95	56,200	3,160,000,000
.00001	.0000000001	100	100,000	10,000,000,000

INSULATING MATERIALS

DIELECTRIC CONSTANTS

Substance	Dielectric Constant* (K)	Substance	Dielectric Constant* (K)
Air	1	Mica	2.5-7.6
Asphalt	2.7-3	Mica, first-quality Indian	6.2-6.7
Bakelite, fabric base	4.6-7	Micanite, non-flexible	7
Bakelite, moulded	4.4-7.8	Mycalex	8.5
Bakelite, paper base	4.5-6.3	Nylon	3.5-3.7
Bakelite, wood filled	4.2-8.1	Paper, varnished	2-3
Casein	6.1-6.8	Paraffin	2.1-2.4
Cambric (varnished)	3.3-5.3	Paraffin oil	2.1-4.7
Celluloid	4-15	Paraffin wax	2-2.5
Celluloid, photographic	6.6-6.7	Paraffined paper	2-3.5
Cellulose acetate	4.5-5.4	Polyethylene	2.5
Cellulose acetate, moulded	3-6	Polystyrene	2.5-2.59
Cellulose acetate, sheet	3-4.5	Polyvinyl chlorido	6-12
Dilectene 100	3.6-3.75	Porcelain	5.5-6.8
Ebonite	2-3.3	Pyrex (glass)	4.5-5
Ebonite, P.O. standard	2.8	Quartz	4.5-5
Empire cloth	2	Rubber	2-3.5
Ethyl cellulose	2-2.9	Shellac	2.9-3.7
Fibre (red)	2.5-4.8	Slate	6-7.4
Fullerboard	3-5	Steatite	6.1
Glass	5.1-9.9	Trolitul	2.2-2.25
Glass, common window	7.9-8.2	Urea-formaldehyde resin	6.6-7.6
Glass, photographic	7.5	Vinylite	3-3.4
Glass, plate	6.5-7.5	Wood (dry)	2.5-7.7
Gutta percha	2.5-4.9	Wood, bakolised	4.5-8
Isolantite	6-6.1	Wood, paraffin-waxed	2.5-8

* Figures are average, but exceptions will be found outside the tolerances quoted. Temperature and frequency also introduce variation; above figures are at 1 mcs. with the exception of cellulose, rubber, shellac and wood.

MANUFACTURERS' BRAND NAMES

Alvar	Polyvinyl Acetal.	Lumarith E.C.	Ethyl Cellulose.
Ameroid	Casein.	Lustron	Polystyrene.
Bakelite	Phenol Formaldehyde.	Makalot	Phenol Formaldehyde.
Bakelite	Urea Formaldehyde.	Melmac	Melamine Formaldehyde.
Bakelite	Polyatryrene.	Micanito	Mica Compounds.
Becolite	Ebonite.	Nixonite	Cellulose Acetate.
Beetle	Urea Formaldehyde.	Nixonoid	Cellulose Nitrate.
Butvar	Polyvinyl Butyral.	Nylon	Supor Polyamide.
Catalin	Phenol Formaldehyde.	Perspex	Methyl-Methacrylate.
Catalin	Melamine Formaldehyde.	Plaskon	Urea Formaldehyde.
Crystalite	Acrylate and Methacrylate.	Pressphan	Laminated Synthetic Resin.
Dexine	Ebonite.	Protectoid	Celluloso Acetate.
Dilectene 100	Aniline Formaldehyde Synthetic Resin.	Prystal	Phenol Formaldehyde.
Distrene	Polystyrene.	Pyralin	Cellulose Nitrate.
Durite	Phenol Formaldehyde.	Resinox	Phenol Formaldehyde.
Durite	Phenolic Furfural.	Steatite	Ceramic Porcelain.
Ebonestos	Synthetic Resin.	Styraflex	Polystyrene.
Erinofort.	Cellulose Acetate.	Styron	Polystyrene.
Erinoid	Casein.	Super Styrex	Polystyrene.
Ethocel	Ethylcellulose.	Temite	Cellulose Acetate.
Fibestos	Cellulose Acetate.	Trolitul	Polystyrene.
Hereaita	Phenol Formaldehyde.	Vinylite A	Polyvinyl Acetate.
Lucite	Methyl Methacrylate.	Vinylite Q	Polyvinyl Chlorido.
		Vinylite X	Polyvinyl Butyral.

ELECTRICAL CONVERSION TABLE

(For Mechanical Conversion Table, see page 69)

MULTIPLE AND SUB-MULTIPLE CONVERSION

To Convert	Into	Multiply by	Conversely, multiply by
Amps	Millamps	1,000	.001
Amps	Microamps	1,000,000	.000001
Cycles	Kilocycles	.001	1,000
Cycles	Megacycles	.000001	1,000,000
Farads	Microfarads	1,000,000	.000001
Farads	Micromicrofarads (pf)	1,000,000,000,000	.000000000001
Henrys	Millihenrys	1,000	.001
Henrys	Microhenrys	1,000,000	.000001
Kilocycles	Megacycles	.001	1,000
Millivolts	Microvolts	1,000	.001
Mhos	Millimhos	1,000	.001
Mhos	Micro-mhos	1,000,000	.000001
Ohms	Micro-ohms	1,000,000	.000001
Ohms	Megohms	.000001	1,000,000
Volts	Millivolts	1,000	.001
Volts	Microvolts	1,000,000	.000001
Volts	Kilovolts	.001	1,000
Watts	Milliwatts	1,000	.001
Watts	Microwatts	1,000,000	.000001
Watts	Kilowatts	.001	1,000

INTER-UNIT CONVERSION

Horsepower	Watts	746	.0034
Horsepower	Force do Cheval	1.0139	.986
Horsepower	B.T.U. per minute	42.4	.02357
Micromicrofarads	Centimetres (capacity)	1.1	.9
Microfarads	Centimetres (capacity)	1,100	.0009
Atmospheres	Inches of Mercurry	29.92	.03342
Atmospheres	Centimetres of Mercury	76	.01316

PREFIXES USED IN RADIO NUMEROLOGY

Prefix	Abbreviation	Definition
centi	c	one-hundredth of
mil or milli	m	one-thousandth of
micro	μ	one-millionth of
micromicro	$\mu\mu$	one-millionth of a millionth of
pica	p	one-millionth of a millionth of
kilo	k	one thousand times
meg or mega	M	one million times

SIGNIFICANCE OF INDICES

Expression	Simple Equivalent	Expression	Simple Equivalent
10^1	10	10^{-1}	.1
10^2	100	10^{-2}	.01
10^3	1,000	10^{-3}	.001
10^4	10,000	10^{-4}	.0001
10^5	100,000	10^{-5}	.00001
10^6	1,000,000	10^{-6}	.000001
10^7	10,000,000	10^{-7}	.0000001
10^8	100,000,000	10^{-8}	.00000001
10^9	1,000,000,000	10^{-9}	.000000001
10^{10}	10,000,000,000	10^{-10}	.0000000001

**TWIST DRILL
SIZES**

**FRACTIONS—DECIMAL—MILLIMETRE
EQUIVALENTS**

Drill Number	Dia. (in.)	Drill Number	Dia. (in.)	Fraction of inch	Decimal of inch	Millimetre	Fraction of inch	Decimal of inch	Millimetre
1	.2280	41	.0960	$\frac{1}{16}$.0156	.397	$\frac{3}{8}$.375	13.097
2	.2210	42	.0935	$\frac{1}{16}$.0133	.394	$\frac{3}{8}$.3513	13.491
3	.2130	43	.0890	$\frac{1}{16}$.0169	1.191	$\frac{3}{8}$.5469	13.891
4	.2090	44	.0860	$\frac{1}{16}$.0025	1.588	$\frac{1}{16}$.5625	14.287
5	.2055	45	.0820	$\frac{1}{16}$.0781	1.985	$\frac{3}{8}$.5781	14.684
6	.2040	46	.0810	$\frac{1}{16}$.0938	2.381	$\frac{3}{8}$.5938	15.081
7	.2010	47	.0785	$\frac{1}{16}$.1094	2.778	$\frac{3}{8}$.6094	15.478
8	.1990	48	.0760	$\frac{1}{16}$.1250	3.175	$\frac{1}{16}$.6250	15.875
9	.1960	49	.0730	$\frac{1}{16}$.1406	3.572	$\frac{3}{8}$.6406	16.272
10	.1935	50	.0700	$\frac{1}{16}$.1563	3.969	$\frac{3}{8}$.6563	16.688
11	.1910	51	.0670	$\frac{1}{16}$.1719	4.366	$\frac{3}{8}$.6719	17.085
12	.1890	52	.0635	$\frac{1}{16}$.1875	4.762	$\frac{1}{16}$.6875	17.462
13	.1850	53	.0595	$\frac{1}{16}$.2031	5.159	$\frac{3}{8}$.7031	17.850
14	.1820	54	.0550	$\frac{1}{16}$.2188	5.556	$\frac{3}{8}$.7188	18.256
15	.1800	55	.0520	$\frac{1}{16}$.2344	5.953	$\frac{3}{8}$.7344	18.653
16	.1770	56	.0465	$\frac{1}{16}$.2500	6.350	$\frac{1}{16}$.7500	19.050
17	.1730	57	.0430	$\frac{1}{16}$.2656	6.747	$\frac{3}{8}$.7656	19.447
18	.1695	58	.0420	$\frac{1}{16}$.2813	7.144	$\frac{3}{8}$.7813	19.843
19	.1660	59	.0410	$\frac{1}{16}$.2969	7.541	$\frac{3}{8}$.7969	20.240
20	.1610	60	.0400	$\frac{1}{16}$.3135	7.937	$\frac{1}{16}$.8125	20.037
21	.1590	61	.0390	$\frac{1}{16}$.3281	8.334	$\frac{3}{8}$.8281	21.034
22	.1570	62	.0380	$\frac{1}{16}$.3438	8.731	$\frac{3}{8}$.8438	21.430
23	.1540	63	.0370	$\frac{1}{16}$.3594	9.128	$\frac{3}{8}$.8594	21.827
24	.1520	64	.0360	$\frac{1}{16}$.3750	9.525	$\frac{1}{16}$.8750	22.224
25	.1495	65	.0350	$\frac{1}{16}$.3906	9.922	$\frac{3}{8}$.8906	22.621
26	.1470	66	.0330	$\frac{1}{16}$.4063	10.319	$\frac{3}{8}$.9063	23.018
27	.1440	67	.0320	$\frac{1}{16}$.4219	10.716	$\frac{3}{8}$.9219	23.415
28	.1405	68	.0310	$\frac{1}{16}$.4375	11.120	$\frac{1}{16}$.9375	23.812
29	.1360	69	.0293	$\frac{1}{16}$.4531	11.509	$\frac{3}{8}$.9531	24.209
30	.1285	70	.0280	$\frac{1}{16}$.4688	11.906	$\frac{3}{8}$.9688	24.606
31	.1200	71	.0260	$\frac{1}{16}$.4844	12.303	$\frac{3}{8}$.9844	25.003
32	.1160	72	.0250	$\frac{1}{16}$.5000	12.700	1	1.0000	25.400
33	.1130	73	.0240						
34	.1110	74	.0225						
35	.1100	75	.0210						
36	.1065	76	.0200						
37	.1040	77	.0180						
38	.1015	78	.0160						
39	.0995	79	.0145						
40	.0980	80	.0135						

WHITWORTH SCREWS

B.S.F. SCREWS

Size (in.)	Turns per inch	Tapping Drill	Size (in.)	Turns per inch	Tapping Drill
$\frac{1}{2}$	40	41	$\frac{7}{8}$	28	$\frac{3}{8}$
$\frac{5}{8}$	24	29	$\frac{1}{2}$	26	$\frac{7}{16}$
$\frac{3}{4}$	20	12	$\frac{5}{8}$	26	$\frac{3}{8}$
$\frac{7}{8}$	18	$\frac{1}{2}$	$\frac{1}{2}$	22	$\frac{3}{8}$
$\frac{9}{16}$	16	$\frac{1}{2}$	$\frac{1}{2}$	20	$\frac{3}{8}$
$\frac{5}{16}$	12	$\frac{1}{2}$	$\frac{1}{2}$	18	$\frac{3}{8}$
$\frac{3}{8}$	11	$\frac{1}{2}$	$\frac{1}{2}$	16	$\frac{3}{8}$
$\frac{1}{4}$	10	$\frac{1}{2}$	$\frac{1}{2}$	16	$\frac{3}{8}$
1	8	$\frac{1}{2}$	$\frac{1}{2}$	14	$\frac{1}{2}$

B.A. SCREWS

B.A. Number	Diameter	Core Diam.	Turns per inch	Clearing Drill	Tapping Drill
0	.2362	.180	25·4	1"	9
1	.2087	.1661	28·2	2	17
2	.185	.1469	31·4	10	24
3	.1614	.1268	34·8	18	20
4	.1417	.1106	38·5	25	32
5	.126	.098	43·0	29	37
6	.1102	.085	47·9	32	43
7	.0984	.0756	52·9	37	46
8	.0866	.0661	59·1	42	50
9	.0748	.0563	65·1	46	53
10	.0669	.0504	72·6	49	54
11	.0591	.0445	81·9	52	56
12	.0512	.0378	90·7	54	60
13	.0472	.0354	101·0	55	63
14	.0394	.0283	110·0	58	68
15	.0354	.0256	121·0	62	70

Tapping Drill sizes are recommended for brass. For soft substances use the next smaller size drill.

WOOD SCREWS

AMERICAN SCREWS (As used in Radio Manufacture)

Gauge Number	Shank Diam.	Clearing Drill	Size Number	Diameter	Tapping Drill
1	.066	48	2-56	.0860	49
2	.080	43	3-48	.0990	44
3	.094	38	4-40	.1120	43
4	.108	32	5-40	.1250	36
5	.122	29	6-32	.1380	33
6	.136	26	8-32	.1640	28
7	.150	21	10-24	.1900	23
8	.164	17	10-32	.1900	20
9	.178	14	12-24	.2160	17
10	.192	9	1-20	.2500	7
11	.206	4			
12	.220	1			
13	.234	1"			

Where danger of splitting is great, use next larger size drill.

MECHANICAL CONVERSION TABLE

To Convert	Into	Multiply by	Conversely, multiply by
Centigrade	Fahrenheit	$(C^\circ \times \frac{9}{5}) + 32$	$(F^\circ - 32) \times \frac{5}{9}$
Cubic inches	Cubic centimetres	16·39	.06102
Gallons (British)	Gallons (U.S.A.)	1·20094	.83268
Grams	Grains	15·432	.0648
Grams	Ounces (a.d.p.)	.03527	28·35
Inches	Centimetres	2·54	.3937
Inches of mercury	Lbs. per sq. in.	.49116	2·0360
Knots	Miles	1·152	.86836
Litres	Pints	1·76	.5882
Metres	Feet	3·2808	.3048
Metres	Yards	1·094	.9144
Metres per minute	Feet per minute	3·281	.3048
Miles per hour	Feet per minute	88	.01136
Miles per hour	Feet per second	1·467	.6818
Sq. inches	Sq. centimetres	6·452	.1560
Watts	B.T.U. per minute	.05688	17·58
Watts	Horse-power	.00134	745·7

READY-WORKED METER SHUNT AND SERIES RESISTANCE TABLE

The table below gives values of shunt resistance for increasing the current range of a milliammeter and the values of series resistance for converting a milliammeter into a voltmeter to cover various ranges. A practical example of this technique is given on page 50. The internal resistance of the milliammeter is a controlling factor. The average 0-1 milliammeter, for example, has an internal resistance of something less than 50 ohms; in drawing up the tables below the author intends that the resistance of the meter shall either be made up to 100 ohms by the use of a suitable series resistance, or, preferably, be made up to 50 ohms with an additional 50 ohms added, the latter being provided with a press-button shorting switch, so that its depression will double the meter reading when desired. By incorrect switching arrangements, serious errors can be introduced by switch contact resistance; attention is therefore drawn to page 12, Vol. III, of *Modern Practical Radio and Television*.

Required Voltage Range (volts)	Series Resistance in Ohms for Milliammeter with range as below				
	1 mA	2 mA	3 mA	5 mA	10 mA
2 . .	*1,900	*900	566	300	*100
3 . .	2,900	1,400	*900	500	200
5 . .	*4,900	2,400	1,560	*900	*400
10 . .	*9,900	*4,900	3,230	*1,900	*900
15 . .	15,000	7,400	*4,900	2,900	1,400
20 . .	*20,000	*9,900	6,560	3,900	*1,900
30 . .	30,000	15,000	*9,900	5,900	2,900
50 . .	*50,000	25,000	16,600	*9,900	*4,900
100 . .	*100,000	*50,000	33,300	*20,000	*9,900
150 . .	150,000	75,000	*50,000	30,000	15,000
200 . .	*200,000	*100,000	66,600	40,000	*20,000
300 . .	300,000	150,000	*100,000	60,000	30,000
500 . .	*500,000	250,000	166,000	*100,000	*50,000
1,000 . .	*1,000,000	*500,000	333,000	*200,000	*100,000

Note.—The above resistance values have been adjusted to convenient values so far as possible consistent with a maximum error not exceeding 1 per cent. Figures marked with asterisk indicate ranges that read conveniently on the meter scale.

SHUNT RESISTANCE IN OHMS FOR MILLIAMMETER WITH TOTAL RESISTANCE OF 100 OHMS

Multiplying Factor		Multiplying Factor		Multiplying Factor		Multiplying Factor	
2	100	10	11·11	60	1·695	400	.251
4	33·33	20	5·26	80	1·266	500	.2
5	25	30	3·45	100	1·01	600	.167
6	20	40	2·56	200*	.503	800	.125
8	14·29	50	2·04	300	.334	1,000	.1

* It is inadvisable to attempt multiplication factors above 100 unless fully conversant with test gear and very low resistance shunts.

COMPARATIVE RESISTANCE

RESISTIVITY, ρ , OF METALS

Material	Relative Resistance	Material	μ Ohms per cu. cm. at 0° C.
Copper . . .	1	Copper, standard . . .	1·589
German silver . . .	11·7-18·5	Copper, hard-drawn . . .	1·6
Eureka . . .	30 average	Eureka . . .	48 average
Nichrome . . .	55 average	Nichrome . . .	112 average
Silver94	Silver . . .	1·47
Silicon bronze . . .	1·5	Aluminium . . .	2·67
Aluminium . . .	1·6	Nickel . . .	12·32
Nickel . . .	4·3	Manganin . . .	42 average
Phosphor bronze . . .	4·4	German silver . . .	21 average
Manganin . . .	27	Iron, annealed . . .	9·07

SULPHURIC ACID TABLE

(Quantities by Volume)

1·350 Sp. Gr.	= 1 part acid to 2·3 parts distilled water
1·300 "	= 1 " " 2·8 " "
1·250 "	= 1 " " 3·6 " "
1·200 "	= 1 " " 5·0 " "
1·150 "	= 1 " " 7·0 " "

Note.—Above table assumes undiluted acid to be 1·840 Sp. Gr. Always add acid to water; the reverse procedure is most dangerous.

ACCUMULATOR DISCHARGE TABLE *

Fully charged	= 1·250 Sp. Gr. at 60° F.
" "	= 1·240 " " 70° F.
" "	= 1·242 " " 80° F.
Half discharged	= 1·180 " " 60° F.
" "	= 1·176 " " 70° F.
" "	= 1·173 " " 80° F.
Discharged	= 1·110 " " 60° F.
" "	= 1·107 " " 70° F.
" "	= 1·105 " " 80° F.

* Varies slightly with different makers.

ESTIMATION OF TEMPERATURE BY COLOUR

Black red	= 900° F. = 500° C.
Deep red	= 1,000° F. = 525° C.
Red	= 1,400° F. = 775° C.
Brilliant red	= 1,550° F. = 850° C.
Vivid orange	= 1,725° F. = 950° C.
Yellow	= 1,825° F. = 1,000° C.
Whitish yellow	= 1,975° F. = 1,100° C.
White	= 2,200° F. = 1,200° C.
Blinding white	= 2,800° F. = 1,500° C.

ESTIMATION OF SOUND INTENSITY

0 db	= Threshold of normal hearing.
10 db	= Very quiet but unmistakable sound.
20 db	= Kettle boiling at 6 ft. away.
30 db	= Sewing machine in adjoining room (door open).
40 db	= Moderate radio set in adjoining room.
50 db	= Traffic noise on main road with closed windows.
75 db	= Very busy suburban main road.
100 db	= Road drill other side of road.
125 db	= Shattering noise, e.g. unsilenced motor-engine fully revved in small confined space; feeling of discomfort begins to over be apparent.
125 db	At noise-levels 130–140 db, pain rather than hearing is experienced.

Note.—The above indications of sound-intensity are obviously wide approximations; they are the result of measurement in a South-west London suburb, using a microphone. The ear, however, will be greatly influenced by the nature of the sound, e.g. a person singing out of tune may give the impression of being louder than a more pleasing sound of equal intensity; also the human ear is not linear to frequency.

CLIMATIC DATA

Miscellaneous information useful when considering overseas requirements or when erecting large aerial masts.

TEMPERATURES

British Isles	= Max. 100° F.	Min.	4° F.
Europe (excluding U.S.S.R.)	=	125° F.	50° F.
Africa	=	136° F.	0° F.
Australia	=	127° F.	19° F.
Asia	=	125° F.	90° F.
North America	=	120° F.	70° F.
New Zealand	=	95° F.	23° F.
South America	=	115° F.	25° F.
U.S.A.	=	134° F.	66° F.
U.S.S.R.	=	110° F.	90° F.

WIND VELOCITY AND PRESSURE

10 m.p.h.	=	23 lb. per sq. ft. on cylindrical surface
20 "	=	8 " " " " "
30 "	=	1·7 " " " " "
50 "	=	4·2 " " " " "
75 "	=	8·7 " " " " "
100 "	=	14·5 " " " " "
125 "	=	21·9 " " " " "
150 "	=	30·9 " " " " "
175 "	=	41·4 " " " " "
200 "	=	53·5 " " " " "
10 "	=	4 lb. per sq. ft. on flat surface
20 "	=	1·3 " " " " "
30 "	=	2·8 " " " " "
50 "	=	9·7 " " " " "
75 "	=	14·5 " " " " "
100 "	=	24·3 " " " " "
125 "	=	36·9 " " " " "
150 "	=	51·9 " " " " "
175 "	=	69·5 " " " " "
200 "	=	89·8 " " " " "

HUMIDITY

Humidity varies so widely in relatively small areas that brief data is meaningless, but in certain parts of the world humidity is as high as 99 per cent. Its effect on radio receivers may be controlled by spraying with nitrocellulose lacquer. For use in non-built-up areas about 2 per cent. of pentachlorophenol may be added as a fungus deterrent.

READY-WORKED NUMERICAL VALUES

π	= 3·1416	2π	= 6·2832
π^2	= 9·8096	$\frac{1}{2\pi}$	= -1·592
$\frac{1}{\pi}$	= -3·183	$\sqrt{2}$	= 1·4142
$\frac{1}{\pi^3}$	= -10·13	$\sqrt{3}$	= 1·7321
$\sqrt{\pi}$	= 1·7725	$\frac{1}{\sqrt{2}}$	= .7071
$\frac{1}{\sqrt{\pi}}$	= -5·042	$\frac{1}{\sqrt{3}}$	= .5773
Area of circle	= πr^2		
Area of triangle	= Base $\times \frac{1}{2}H$		
Area of cylinder	= $2\pi rH$		
Volume of sphere	= $\frac{4}{3}\pi r^3$		
Volume of cylinder	= $\pi r^2 H$		
Volume of cone	= $\frac{\pi r^2 H}{3}$		

VARIATIONS IN WORLD TIME

Local time is given in considerable detail below owing to its great value when identifying foreign stations. Up-to-date information of this kind and up-to-date frequency changes are given in the *Short-Wave News*, published by The Amalgamated Short-Wave Press Ltd., 57 Maida Vale, W.9.

<i>Country</i>	<i>To G.M.T. ADD</i>	<i>From G.M.T. TAKE</i>	<i>Country</i>	<i>To G.M.T. ADD</i>	<i>From G.M.T. TAKE</i>
	<i>Hrs. Mins.</i>	<i>Hrs. Mins.</i>		<i>Hrs. Mins.</i>	<i>Hrs. Mins.</i>
Aden	3 00	—	Kenya	2 30	—
Afghanistan	4 00	—	Korea	9 00	—
Albania	1 00	—	Latvia	2 00	—
Algeria	No change	—	Liberia	—	44
Argentine	—	4 00	Libya	1 00	—
Australia, N.S.W., Victoria	10 00	—	Lithuania	1 00	—
" Queensland	9 30	—	Madagascar	3 00	—
" Western	8 00	—	Malay States	7 20	—
Bahamas	—	5 00	Malta	1 00	—
Barbados	—	4 00	Manchukuo	8 00	—
Bechuanaland	2 00	—	Mexico	—	6 00
Belgian Congo (Loopoldville)	1 00	—	Mongolia	7 00	—
" (Stanleyville)	2 00	—	Mozambique	2 00	—
Belgium	No change	—	Netherlands	—	20
Bermuda	—	4 00	Newfoundland	—	3 30
Bolivia	—	4 33	New Guinea (Dutch)	9 00	—
Borneo, North	8 00	—	" (British)	10 00	—
" Dutch	7 30	—	New Zealand	11 30	—
Brazil	—	3 00	Nigeria	1 00	—
British Guiana	—	3 45	Norway	1 00	—
British Honduras	—	6 00	Palestino	2 00	—
Bulgaria	2 00	—	Panama	—	5 00
Burma	6 30	—	Paraguay	—	4 00
Canada:			Peru	—	5 00
New Brunswick, Nova Scotia	—	4 00	Poland	1 00	—
Quebec, Ontario	—	5 00	Portugal	—	No change
Manitoba	—	6 00	Rhodesia, North	2 00	—
Alberta, Saskatchewan	—	7 00	" South	2 00	—
British Columbia	—	8 00	Rumania	2 00	—
Yukon Territory	—	9 00	Saudi Arabia	3 00	—
Ceylon	5 00	—	Senegal	—	1 00
Chile	—	4 00	Sierra Leone	—	1 00
China, Eastern	8 00	—	Solomon Islands	10 00	—
" Central and Western	7 00	—	Somaliland (Fr. and Brit.)	3 00	—
Colombian Republic	—	5 00	South-West Africa	2 00	—
Costa Rica	—	5 00	Spain	—	No change
Cuba	—	5 00	Sudan	2 00	—
Cyprus	2 00	—	Sweden	1 00	—
Czechoslovakia	1 00	—	Switzerland	1 00	—
Denmark	1 00	—	Syria	2 00	—
Ecuador, Guayaquil	—	5 19	Tanganyika	3 00	—
" all other parts	—	5 14	Thailand	7 00	—
Egypt	2 00	—	Transjordan	2 00	—
El Salvador	—	6 00	Trinidad and Tobago	—	4 00
Estonia	2 00	—	Tunisia	1 00	—
Ethiopia	3 00	—	Turkey	2 00	—
Finland	2 00	—	Union of South Africa	2 00	—
France	No change	—	U.S.S.R., European	3 00	—
French Cameroons	1 00	—	" Ukraine (Kiev)	4 00	—
" Equatorial Africa	1 00	—	" Ural Area (Sverdlovsk)	5 00	—
" Guyana	—	4 00	" Usbek (Tashkent)	6 00	—
" Indo-China	7 00	—	" Siberian Area (Novosibirsk)	7 00	—
Gambia	—	1 00	" Siberian Area (Irkutsk)	8 00	—
Germany	1 00	—	" Asiatic Repb. (Chita, etc.)	9 00	—
Gold Coast	No change	—	" Asiatic Repb. (Vladivostok)	10 00	—
Greece	2 00	—	U.S.A., Atlantic Coast	—	5 00
Greenland	—	3 00	" Central States	—	6 00
Hawaiian Islands	—	9 30	" Mid-Western States	—	7 00
Honduras	—	6 00	" Pacific Coast	—	8 00
Hungary	1 00	—	Uruguay	—	3 30
India	5 30	—	Venezuela	—	4 30
Iran	3 00	—	Yugoslavia	1 00	—
Iraq	3 00	—			
Italy	1 00	—			
Jamaica	—	5 00			
Japan	9 00	—			
Java	7 30	—			

SYMBOLS

Note.—Symbols used are those in daily use and are not necessarily idealistic.

GENERAL SYMBOLS

Admittance	Y, y	Magnetic field.	H
Ampero	A	Magnetic flux	Φ
Amplification factor	μ, m	Magnetic flux density	B
Angular velocity ($2\pi f$)	ω	Magneto motive force	F
Anode A.C. resistance (impedance)	R_a	Magnification of tuned circuit	Q
Anode current	I_a	Mutual inductance	M
Anode current inductance	L_a	Number of plates, turns, etc.	N, n
Anode voltage	V_a	Ohm	Ω
Capacity	C	Period	T
Conductance	G, g	Permeability	μ
Conductivity	γ	Phase displacement or angle	ϕ
Current (instantaneous)	i	Power output	P_o
Current (R.M.S.)	I	Reactance	X
Dielectric constant	K	Reactance, capacitive	X_c
E.M.F.	E	Reactance, inductive	X_L
E.M.F. (instantaneous)	e	Reluctivity	v
Energy	W	Resistance	R
Farad	F	Resistance, H.F.	r
Frequency	f	Resistivity	ρ
Grid circuit inductance	L_g	Susceptance	b
Grid current	I_g	Time	t
Grid voltage	V_g	Velocity	v
Henry	H	Volt	V
Impedance	Z	Watt	W
Inductance	L	Wavelength	λ

VALVE SYMBOLS *

Amplification factor	μ, m	Conversion conductance	g_c
Anode current	I_a	Filament current	I_f
Anode impedance	R_a	Filament voltage	V_f
Anode voltage	V_a	Grid current	I_g
Capacity, anode cathode	C_{ac}	Grid input impedance	Z_g
Capacity, anode output	C_a	Grid voltage	V_g
Capacity, grid anodo	C_{ga}	Optimum load	R_o
Capacity, grid cathode	C_{gc}	Screen current	I_{gs}
Capacity, grid input	C_g	Screen voltage	V_{gs}
Cathode current	I_c	Slope	gm
Cathode voltage	V_c	Transconductance (U.S.A.)	gm

* A new system of valve symbolisation has been introduced by the B.V.A. and will doubtless pass into general usage in due course; that shown has been in use for many years and, with slight variation, will be found in most text-books.

MATHEMATICAL SYMBOLS

Is equal to	$=$	Less than	$<$
Is not equal to	\neq	Not less than	\geq
Is approx. equal to	\approx	The sum of	Σ
Is the same as	\equiv	General symbol of an angle	θ
The difference between	\sim	A small difference	δ
Varies as ; is proportional to	\propto	Anglo	\wedge
Greater than	$>$	An unknown quantity	x
Not greater than	$\not>$	Angular velocity ; $2\pi f$	ω

C.G.S. MECHANICAL SYMBOLS

Length (cm.)	l	Force (dyne)	F
Mass (gm.)	m	Work (erg)	W
Time (sec.)	t	Energy (erg)	W
Surface (sq. cm.)	A	Power (erg per sec.)	P
Volume (cu. cm.)	V	Pressure (dynes per sq. cm.)	P
Velocity (cm. per sec.)	v	Angle (radian)	ϕ
Acceleration (cm. per sec. ²)	a	Angular velocity (radians per sec.)	ω

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VALVE BASE CONNECTIONS

Every care has been taken in compiling the information given below; it will be realised that the author is dependent, in many cases, upon information received from various sources; errors may therefore exist and no liability can be accepted.

A solid arc drawn through the edge of a base pin indicates an internal shield connection; a similar arc drawn dotted illustrates a metallising connection.

FOR CORRESPONDING VALVE BASE DIAGRAMS, SEE PAGES 80-85.

<u>BRIMAR</u>	6A8EG	O33	G.P. and Output Pentodes and Tetrodes with and without Diodes	Class B and Q.P.P. BATTERY
S.G. and H.F. Pens.	6A8G	O33		
BATTERY	6F7	UX105		
1A4E	UX7	6F7E	UX105	
1LN5E	L5	6K8G	O36	
1N5EG	O14	6L7G	O31	
15E	UX29	6P8G	O35	
32E	UX7	7A8E	L11	
34E	UX7	7B8	L10	
S.G. and H.F. Pens.	12K8GT	O36	1A5G O23†	
MAINS	15A2	B132	1C5G O23†	
6B8SG	O28	15D1	ILA4E L8	
6C6	UX53	15D2	1Q5G O21†	
6D6	UX53	20A1	3Q5G O21	
6F7B	UX105	20D2	2101 UX27	
6J7G	O17		PENAI B59	
6K7EG	O16		† Ignore filament centre	
6K7G	O16		tap	
6U7G	O16			
7A7	L4	1H5G	G.P. and Output Pentodes and Tetrodes with and without Diodes	Tuning Indicators
7A7E	L4	1LH4	BATTERY	6U5/6G5 UX60
7B7E	L4	27	MAINs	
7C7E	L4	30	6AG6G O26	
7H7	L4	45	6B7 UX100	
8A1	B61	2102	6B7E UX100	
8A1	B113		5R4GY O2	
8D2	B120		5U4G O2	
9A1	B61		5V4G O5	
9A1	B113		5X4G O3	
9D2	B120		5Y3G O2	
12J7GT	O16	4D1	6F6EG O25	
12K7GT	O16	6A3	6F6G O26	
24A	UX29	6B4G	5Z3 UX3	
24E	UX29	6B6G	6K6G O26	
36	UX29	6C5G	6L6G O22	
36E	UX29	6F5G	6V6G O22	
39/44	UX29	6H6G	7A2 B117	
39/44E	UX29	6J5G	7A3 B117	
77	UX53	6K5G	7B7E L7	
77E	UX53	6Q7G	7C5 L6	
78	UX53	6R7G	7D3 B117	
78E	UX53	7B6	7D6 B117	
Frequency Changers	10D1	B105	7D8 B117	
BATTERY	11A2	7C6	7D5 B117	
1A6	UX56	11D3	7D6 B117	
1A7EG	O32	11D5	7D8 B117	
1A7GT	O32	12Q7GT	12A6 O22	
1C6	UX56	37	12A7 O22	
1LA6E	L9	71A	12C8GT O28	
Frequency Changers	75	UX52	12C8GT O28	
MAINS	76	UX26	12C8GT O28	
6A7	UX101	85	12C8GT O28	
6A7E	UX101	HLA2	12C8GT O28	
		B54	12C8GT O28	
		PA1	12C8GT O28	
		B54	12C8GT O28	

<u>COSSOR</u>
S.G. and H.F. Pens.
BATTERY

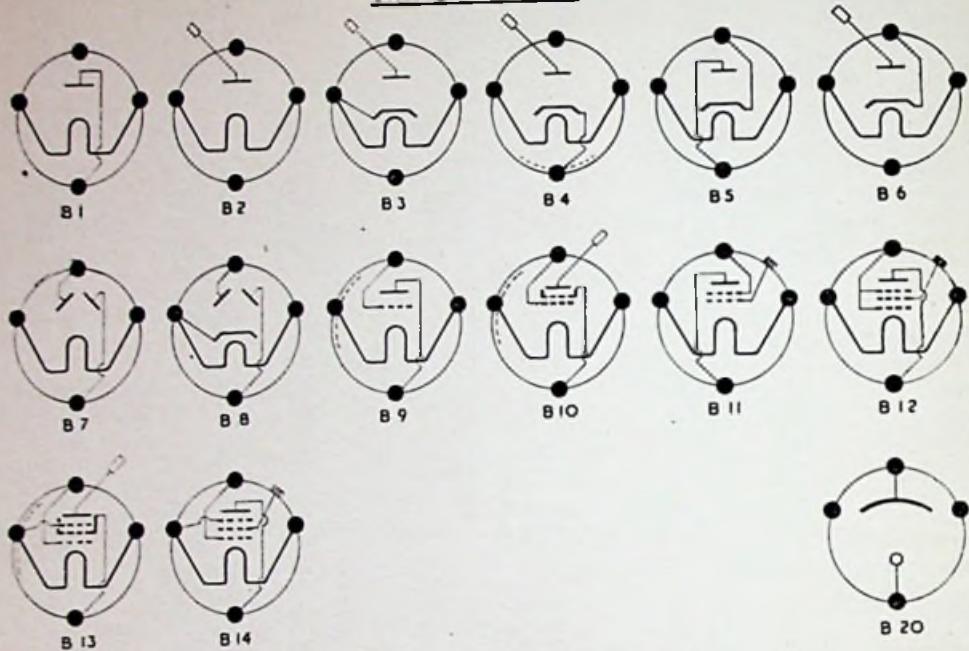
COSSOR—continued		Frequency Changers	Output Triodes	Double Valves			
S.G. and H.F. Pens.		MAINS	MAINS	MAINS			
BATTERY							
15	UX29	4THA 6A7E 6A8EG	B127 UX101 O33	2P 2XP 4XP	B9 B9 B9	6N7G 79	O37 UX58
32E	UX7	6F7B	UX105	6A3	UX5		
34E	UX7	6K8G	O36	6B4G	O8		
210SPT	B13	13PGA	B132	6J5G	O9		
210SPT	B110	41MDG	B60	41MP	B54		
210VPA	B110	41MPG	B132	41MXP	B54		
210VPT	B13	41STH	B127	45	UX5	Tuning Indicators and Miscellaneous	
210VPT	B110	202MPG	B132	402P	B103	4TP 41ME	B136 SC50
215SG	B10	202STH	B127	DP	B54	S130	B20
220IPT	B112	302THA	B127				
220SG	B10	OM8	O33				
220VS	B10	OM10	O35				
220VSG	B10						
S.G. and H.F. Pens.							
MAINs							
4TPB	B114	1H5G 30	O12 UX5	Diodes, Triodes and Diode Triodes	BATTERY	G.P. and Output Pentodes and Tetrodes with and without Diodes	Rectifiers
4TSA	B140	210DDT	B56	1C5G	O23†	4/100BU 5X4G 5Y4G	B7 O3 O3
4TSP	B113	210DET	B9	220HPT	B14	5Z3	UX3
6J7G	O16	210HF	B9	220HPT	B62	6ZY5G	O6
6K7G	O16	210HL	B9	230PT	B58	25RE	UX50
13SPA	B114	210LF	B9	230PT	B14	25Z5	UX50
13VPA	B114	210RC	B9	2101	UX27	35RE	UX50
24E	UX29	220DD	B51	† Ignore filament centre tap		40SUA	B50
36E	UX29	2102	UX51			43IU	B8
39/44E	UX29					44IU	B8
41MSG	B57					45IU	B8
41MPT	B113					80	UX3
41MTS	B139					84	UX25
42MPT	B113					225DU	B103
42PTB	B114	6H6G	O7	Diodes, Triodes and Diode Triodes	MAINs	405BU	B7
42SPT	B113	6K5G	O10	6B7E	UX100	442BU	B7
77E	UX53	6R7G	O13	6B8EG	O28	460BU	B7
78E	UX53	6Q7G	O13	6F6EG	O26	506BU	B7
202SPB	B114	13DHA	B106	6K6G	O26		
202VP	B113	27	UX26	18E	UX55		
202VPB	B114	37	UX26	25A6G	O26		
DS/PEN	B61	41MH	B54	38E	UX28		
DVSG	B57	41MHF	B54	40PPA	B117		
DVS/PEN	B61	41MHL	B54	41E	UX55		
MSG/H.A	B57	41MLF	B54	42E	UX55		
MSG/L.A	B57	41MTA	B54	42MP/PEN	B117		
MS/PEN	B61	41MTB	B54	42OT	B107		
MS/PEN	B113	41MTL	B54	42OTDD	B109		
MS/PEN/B	B114	75	UX52	43E	UX55		
MVSG	B57	76	UX26	47E	UX27		
MVS/PEN	B61	85	UX52	402OT	B108		
MVS/PEN/B	B114	202DDT	B106	402PEN	B119		
MVS/PEN	B113	DD4	B51	402PEN/A	B119		
OM5	O16	DDL4	B51	807	UX30		
OM6	O16	DDT	B106	2151	UX54		
		DDT16	B106	DD/PEN	B121		
		DHL	B54	402PEN	B117		
		GDT4	B66	MP/PEN	B64		
		GDT4B	B66	OM9	O27		
		OM3	O7	PT10	B117		
		OM4	O13	PT41	B62		
Frequency Changers				Double Valves	MAINs		
BATTERY				BATTERY			
1A6E	UX56			19	UX59	10F9	NBS20
1A7G	O32			220B	B137	AC/S2PEN	B113
1A7VG	O32			240B	B137	AC/SP1	B113
1C6E	UX56			240QP	B138	AC/SP3	B114
210DG	B59	215P	B9	2103	UX106	AC/VP1	B113
210PG	B131	220P	B9			AC/VP2	B114
210PGA	B131	220PA	B9			SP41	MO12
210SPG	B131	230XP	B9			SP42	MO12
220TH	B130					VP41	MO12
						VP133	MO12
						VP1321	B113
						VP1322	B114

<u>MAZDA—continued</u>		T31	B53	<u>FERRANTI</u>	<u>MULLARD</u>	
<u>Frequency Changers</u>		T41	MO7	Battery Types	<u>S.G. and H.F. Pens.</u>	
BATTERY		V312	B55	H2D	B56	<u>BATTERY</u>
FC141		V914	B51	HP2	B137	1A4 UX7
TP22				L2	B9	IN5G O14
TP23				VHT2	B130	15 UX29
TP25				VHT2A	B130	32 UX7
TP26				VPT2	B110	34 UX7
Output Tetrodes and Pentodes with and without Diodes				VS2	B10	DF1 SC13
BATTERY						DF33 O15
<u>Frequency Changers</u>		PEN24	MO17			KF35 O14
MAINS		PEN25	MO17			PM12 B10
10C1		PEN141	MO17	S.G. and H.F. Pens.		PM12A B10
AC/TH1		PEN220	B62	MAINS		PM12M B10
AC/TP		PEN231	B62	6J7G	O17	SP2 B110
TH41				6K7G	O16	VP2 B110
TH233				SPT4A	B113	VP2B B125
TH2320				SPTS	B113	
TH2321				VPT4	B61	
TP2620				VPT4B	B113	
Output Tetrodes and Pentodes with and without Diodes				VPTA	B113	
MAINS		6P25	O22	VPTS	B113	2A6 UX52
10P13						2B7 UX100
10P14						6C8 UX53
AC/PEN						6D6 UX53
AC2/PEN						6F7 UX105
BATTERY		AC2/PENDD	B122	Frequency Changers		6J7G O16
DD207		B7	AC4/PEN	MAINS		6K7G O16
H141D		MO8	AC5/PEN	6A8G	O33	6S7G O16
HL2		B9	AC5/PENDD	6K8G	O36	6U7G O16
HL21/DD		B56	AC6/PEN	VHT4	B129	24A UX29
HL22		MO5	AC6/PEN	VHTA	B129	35 UX30
HL23		MO5	PEN44	VHTS	B129	36 UX29
HL23/DD		MO9	PEN45	MO14		39/44 UX29
L2		B9	PEN45DD	MO16		58 UX53
L21/DD		B56	PEN46	MO15		77 UX53
L22/DD		MO9	PEN383	MO14		78 UX53
P20		B9	PEN453DD	MO16		EF5 SC14
P220A		B9	PEN3520	B117		EF6 SC14
PENDD4020			PENDD4020	B122		EF8 SC20
PENDD4021				B109		EF9 SC14
<u>Diodes, Triodes and Diode Triodes</u>						
MAINS				6Q7G	O13	
6D1		Misc. 1	PD220	D4	B54	
GL18		NBS10	PD220A	DA	B105	
10LD11		NBS15	QP25	H4D	B106	
AC2HL		B54	QP230	HAD	B106	
ACHL		B54	QP240	HSD	B106	
AC/HLDD		B106		LP4	B9	MM4V B57
AC/P		B54		ZD	B51	S4VA B57
AC/P1		B54				S4VB B57
AC/P4		B53				SP4 B61
DD41		MO4	ME41	MO50		SP4B B113
DD101		MO4	ME91	6B8G	O28	SP13B C114
DD620		B51	ME920	6F6G	O26	VP4 B113
HL41		MO7		6V6G	O22	VP4B B114
HL41DD		MO10		PTA	B117	VP13A SC14
HL42DD		MO10		PT4	B117	VP13C B114
HL133		MO6	MU2	PT4D	B109	
HL133DD		MO10	U21	PTSD	B122	
HL1320		B105	U22	PTZ	B119	
HLDD/1320		B106	U201			
P41		MO7	U403			
PP3/250		B9	U404	MO1		
PA20		B9	U4020	NBS1		
PA40		B9	UU4	5Y3G	O2	
PP5/400		B9	UU5	R4	B7	
PP3521		B104	UU6	R8	R4A	
T11		B53	UU7	MO3	R7	
T21		B53	UU8	MO3	RS	
Rectifiers				RA	B51	
Rectifiers				RZ	B50	
Rectifiers						KK32 O32
BATTERY						TH2 B126
<u>Frequency Changers</u>						
BATTERY						
1A6						
1A7G						
1C6						
1C7G						
DK1						
DK32						
FC2						
FC2A						
TH2						

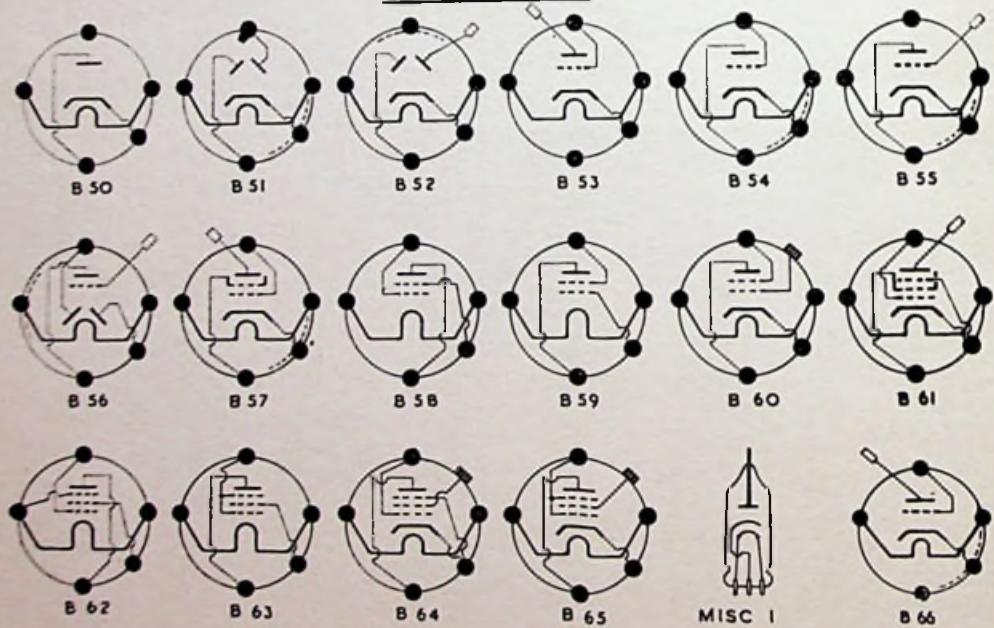
MULLARD—continued		Rectifiers and Doublers	
Frequency Changers			
MAINS			
2A7	UX101	EB4	SC5
6A7	UX101	EBC3	SC11
6A8G	O33	EBC33	O13
6D8G	O33	HL13	SC8
6L7G	O31	HL13C	B105
ECH2	SC24	TDD4	B106
ECH3	SC24	TDD13C	B106
ECH33	O35	† Ignore second anode and top cap	
EK2	SC22	Output Triodes	
EK3	SC22	PM2	B9
FC4	B133	PM2A	B9
FC13	SC22	PM202	B9
FC13C	B133	PM252	B9
TH4	B129	BATTERY	
TH4B	B129	CL4	SC16
TH21C	BL28	CL6	SC16
TH22C	BL28	CL33	O26
TH30C	B129	EBL1	SC18
Diodes, Triodes and Diode Triodes		EBL31	SC18
BATTERY		EL2	6X5G
1H4G	O8	EL3	12A7
1H5G	O12	EL6	25Y5
2D2	B51	EL32	25Z5
30	UX5	EL33	25Z6G
2102	UX51	EL2	O26
DAC1	SC9	EL3	SC16
DAC32	O12	EL6	SC17
KBC32	O13*	EL32	SC17
PM1A	B9	EL33	O27
PM1HF	B9	EL2	AZ3
PM1HL	B9	EL3	AZ31
PM2DL	B9	EL36	O26
PM2DX	B9	PEN4DD	B123
PM2HL	B9	PEN4VA	B64
TDD2	B56	PEN4VA	B117
TDD2A	B56	PEN4VB	B117
* Ignore cathode		PEN4VB	CY1
Diodes, Triodes and Diode Triodes		PEN36C	CY2
MAINS		PEN40DD	B117
2D4A	B51	PEN428	CY31
2D4B	B101	PENA4	B117
2D13	SC27	PENB4	B117
2D13A	SC27†	PM24	CY32
2D13C	B51	PM24	DW2
6C5G	O9	PM24A	DW3
6F5G	O11	PM24B	B7
6H6G	O7	PM24E	B7
6J5G	O9	PM24M	DW4/350
6K5G	O10	Pentodes and Diode Pentodes	
6L5G	O9	1C5G	B123
6Q7G	O13	IF4	B117
6R7G	O13	IF5G	B117
27	UX26	IF7G	B117
56	UX26	BATTERY	
75	UX52	O23†	SC16
76	UX26	UX27	SC16
85	UX52	UX27	SC16
164V	B54	DL1	SC15
244V	B54	DL2	SC15
354V	B54	DL33	O23
904V	B54	KL35	O23
EA50	Misc.1	PM22	B14
EAB1	SC6	PM22A	B62
Tetrodes and Pentodes with and without Diodes		PM22A	B137
MAINS		QP22A	B137
75	UX52	QP22B	B138
76	UX26	Double Valves	
85	UX52	19	UX59
164V	B54	2103	UX106
244V	B54	PM2B	B137
354V	B54	PM2BA	B137
904V	B54	QP22A	B202
EA50	Misc.1	QP22B	B138
EAB1	SC6	BATTERY	
* Ignore cathode		6A6	UX102
Diodes, Triodes and Diode Triodes		6B5	UX57
MAINS		6N6G	O39
2D4A	B51	6N7G	O37
2D4B	B101	79	UX58
2D13	SC27	Double Valves	
2D13A	SC27†	S21	B10
2D13C	B51	S22	B10
6C5G	O9	S23	B10
6F5G	O11	S24	B10
6H6G	O7	MAINS	
6J5G	O9	6A6	UX102
6K5G	O10	6B5	V52
6L5G	O9	6N6G	VP21
6Q7G	O13	* Ignore cathode	B110
6R7G	O13	6N7G	B10
27	UX26	79	VS24
56	UX26	Tetrodes and Pentodes with and without Diodes	
75	UX52	EM1	W21
76	UX26	EM3	Z14
85	UX52	EM4	O14
164V	B54	6U5/6G5	Z21
244V	B54	EFM1	B13
354V	B54	EM1	O16
904V	B54	EM3	O16
EA50	Misc.1	EM4	O16
EAB1	SC6	TV4	O27
Tuning Indicators		TV4A	O27
MAINS		TV6	O27
75	UX52	TV6	KTW61
76	UX26	UX60	KTW61M
85	UX52	UX60	KTW63
164V	B54	EFM1	KTW73M
244V	B54	EM1	O27
354V	B54	EM3	O27
904V	B54	EM4	KTZ41
EA50	Misc.1	6F6G	KTZ63
EAB1	SC6	O22	KTZ73M
S.G. and H.F. Pens.		TV6	MS4
S.G. and H.F. Pens.		SC50	B116
S.G. and H.F. Pens.		SC50	O27
S.G. and H.F. Pens.		SC50	O27
S.G. and H.F. Pens.		SC50	O27
S.G. and H.F. Pens.		SC50	KTZ41
S.G. and H.F. Pens.		SC50	KTZ63
S.G. and H.F. Pens.		SC50	KTZ73M
S.G. and H.F. Pens.		SC50	O27
S.G. and H.F. Pens.		SC50	MS4
S.G. and H.F. Pens.		SC50	B57

MARCONI		Diodes, Triodes and Diode Triodes		Double Valves		H.F. Pentodes and Frequency Changers	
OSRAM—continued		MAINS		BATTERY AND MAINS		MAINS	
S.G. and H.F. Pens.		D 41	B 51	B21	B137	A36A	B127
	MAINS	D 42	B 5	BL62	O38	A36C	B129
MS4B	B57	D 43	B 6	DET19	UX103	A50A	B113
MSP4	B57	D 63	O 7	QP21	B138	A50B	B114
MSP4	B113	DA30	B 9			A50M	B113
MSP41	B57	DA41	UX5			A50N	B61
MSP41	B113	DH30	B106			A50N	B113
VMP4	B57	DH42	B106			A50P	B114
VMP4	B113	DH03	O13	Tuning Indicators		A80A	B133
VMP4G	B113	DH73	O13	Y61	O50	C36A	B127
VMS4	B57	DL63	O13	Y62	O50	C36C	B129
VMS4B	B57	H63	O11	Y63	O50	C50B	B114
W42	B120	L63	O9	Y64	O50	C50N	B114
Z62	O16	MH4	B54	Y73	O50	C80B	B133
		MH40	B54			ECH3	SC24
		MH41	B54			EF9	SC14
		MHD4	B106			EF39	O16
Frequency Changers		MHL4	B54				
	BATTERY	ML4	B54				
X14	O32	PX4	B9	Rectifiers			
X21	B134	PX25	B9	GU5	B2	Diodes, Triodes	
X22	B134	PX25A	B9	GU50	B2	and Diode Triodes	
X23	B126			MU12/14	B8	MAINS	
X24	B126			U10	B7	A20B	B51
				U12/14	B7	A23A	B106
				U16	B2	A30B	B54
				U17	B2	A30D	B54
				U18/20	B7	C20C	B51
Frequency Changers		BATTERY		U30	B102	C23B	B106
	MAINS	KT2	B62	U31	O1	C30B	B105
MX40	B132	KT21	B62	U50	O2	EB34	O7
X30/32	B132	KT24	B62	U52	O2	EBC3	SC11
X31	B127	N14	O23†	U70	O6	EBC33	O13
X41	B127	PT2	B62	U71	O1	S30C	B9
X42	B132			U134	O7	S30D	B9
X61M	O35	† Ignore filament centre tap					
X62	O35						
X63	O33						
X63M	O33						
X65	O35						
X71M	O35						
X73M	O33						
X75	O35						
Triodes and Diode Triodes		Output Pentodes and Tetrodes with and without Diodes		EVER READY		Output Pentodes and Diode Pentodes	
	BATTERY	MAINs		All Types		MAINS	
H2	B9	KT32	O22	K23A	B56	A27D	B123
HD14	O12	KT63	O22	K23B	B56	A70B	B117
HD22	B56	KT66	O22	K30A	B9	A70D	B117
HD23	B56	KT72	O22	K30C	B9	A70E	B117
HD24	B56	KT73	O22	K30D	B9	C70D	B117
HL2	B9	MKT4	B64	K30E	B9	EBL1	SC18
HL21	B9	MKT4	B107	K33B	B137	EBL31	O28
L21	B9	MPT4	B64	K40B	B10	EL3	SC17
L210	B9	N30	B117	K40N	B10	EL32	O27
LP2	B9	N30G	B117	K50N	B125	EL33	O26
P2	B9	N41	B117	K70B	B62	Rectifiers	
P215	B9	N42	B117	K70D	B62	A11C	B8
				K77A	B202	A11D	B8
				K80A	B134	AZ1	SC2
				K80B	B135	AZ31	O2
						C10B	B50
						CY31	O1
						S11A	B7
						S11D	B7

BRITISH 4 PIN

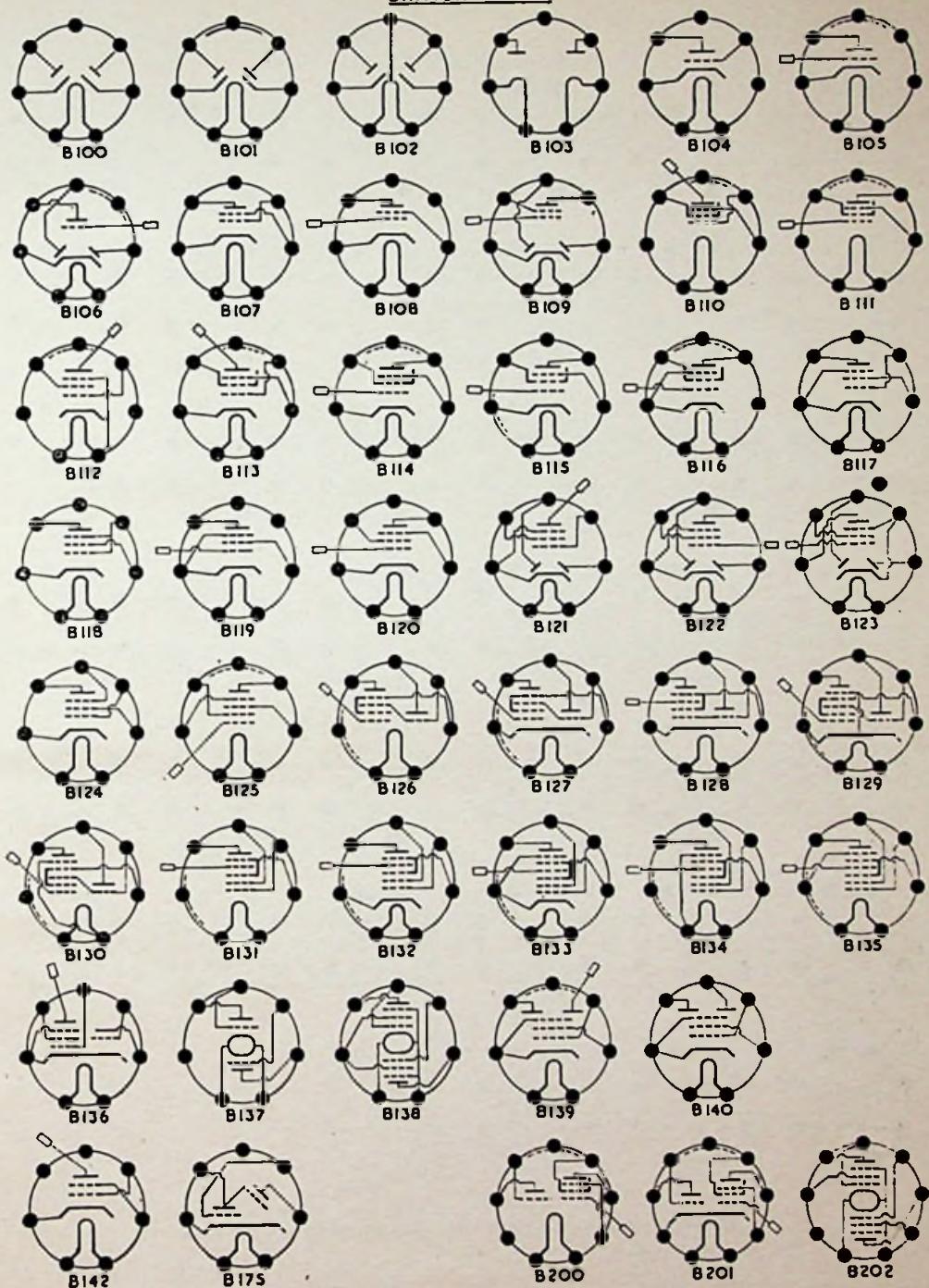


BRITISH 5 PIN



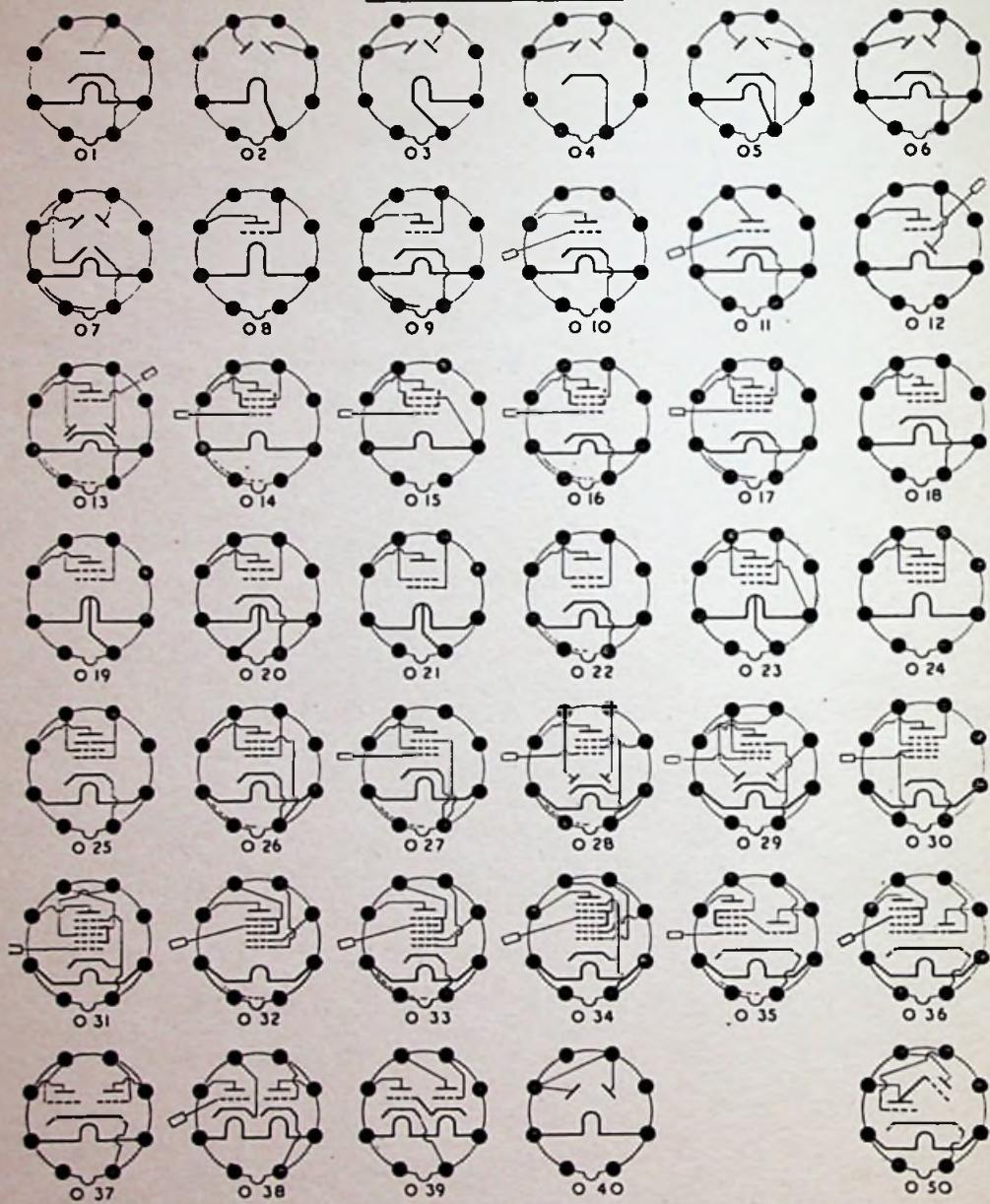
Note.—All connections are as viewed from underside of valve base.

BRITISH 7 PIN

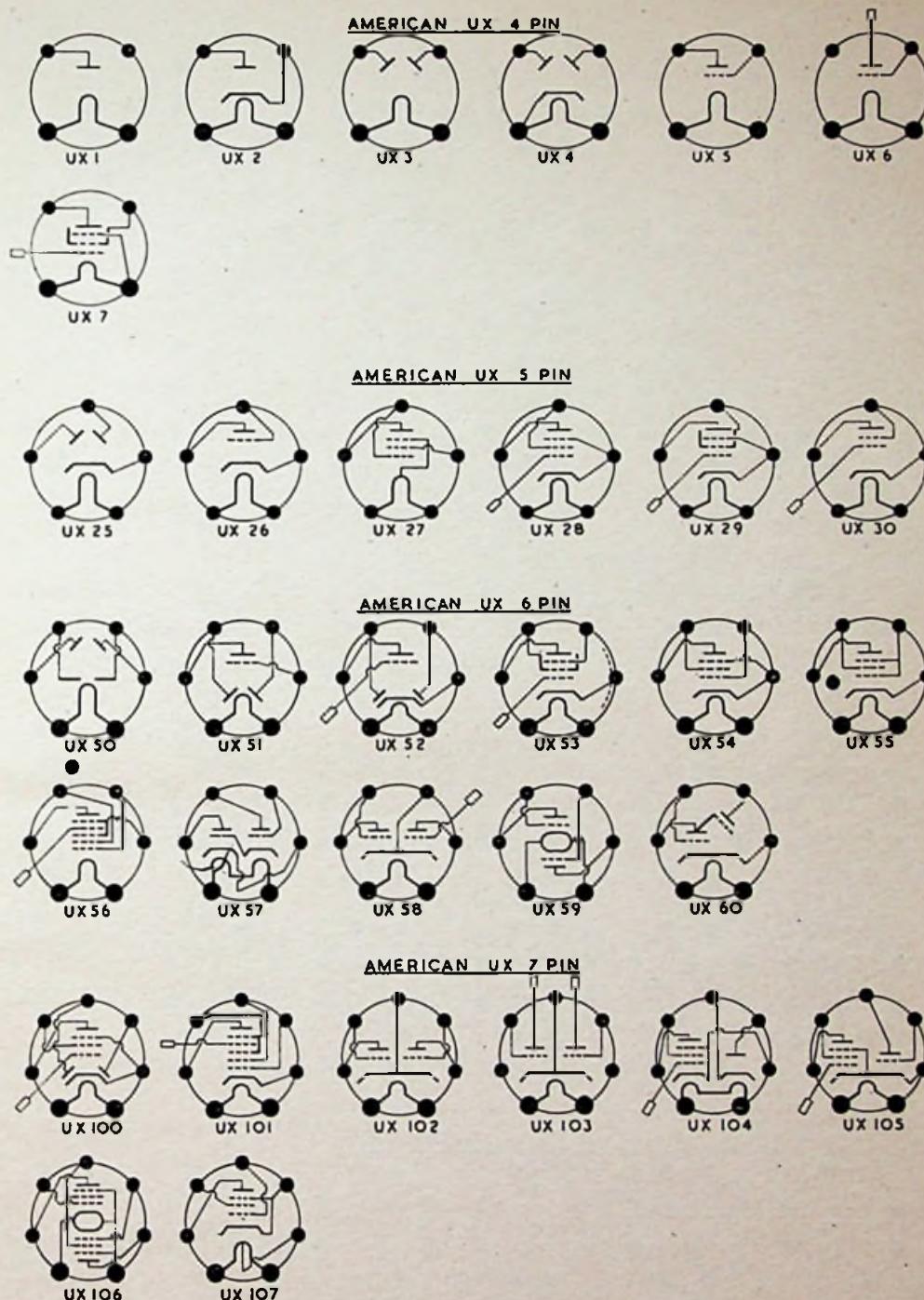


Note.—All connections are as viewed from underside of valve base.

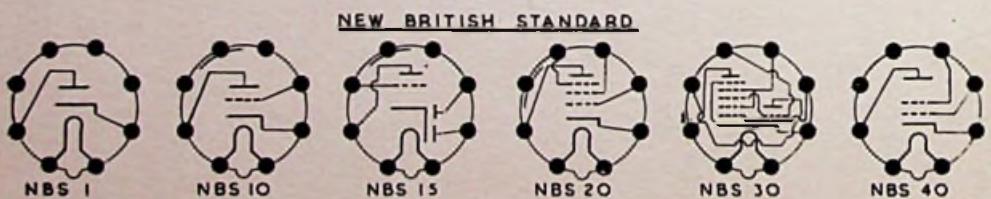
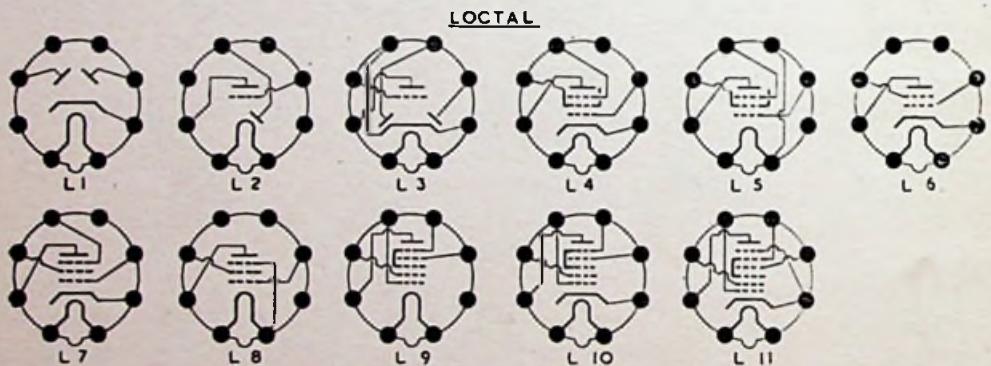
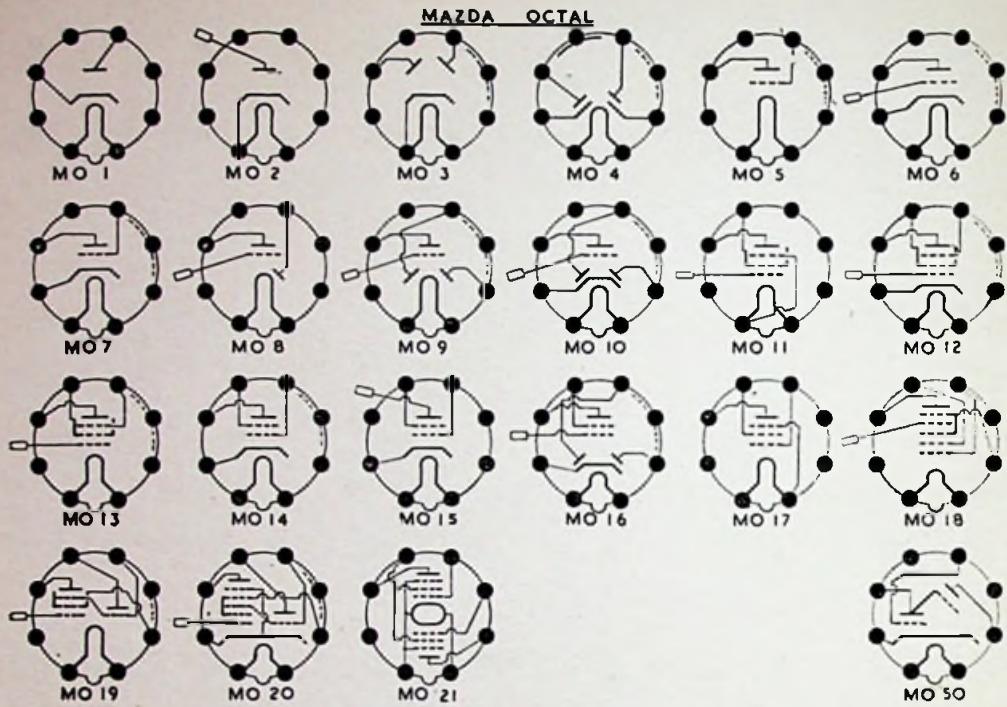
INTERNATIONAL OCTAL



Note.—All connections are as viewed from underside of valve base.

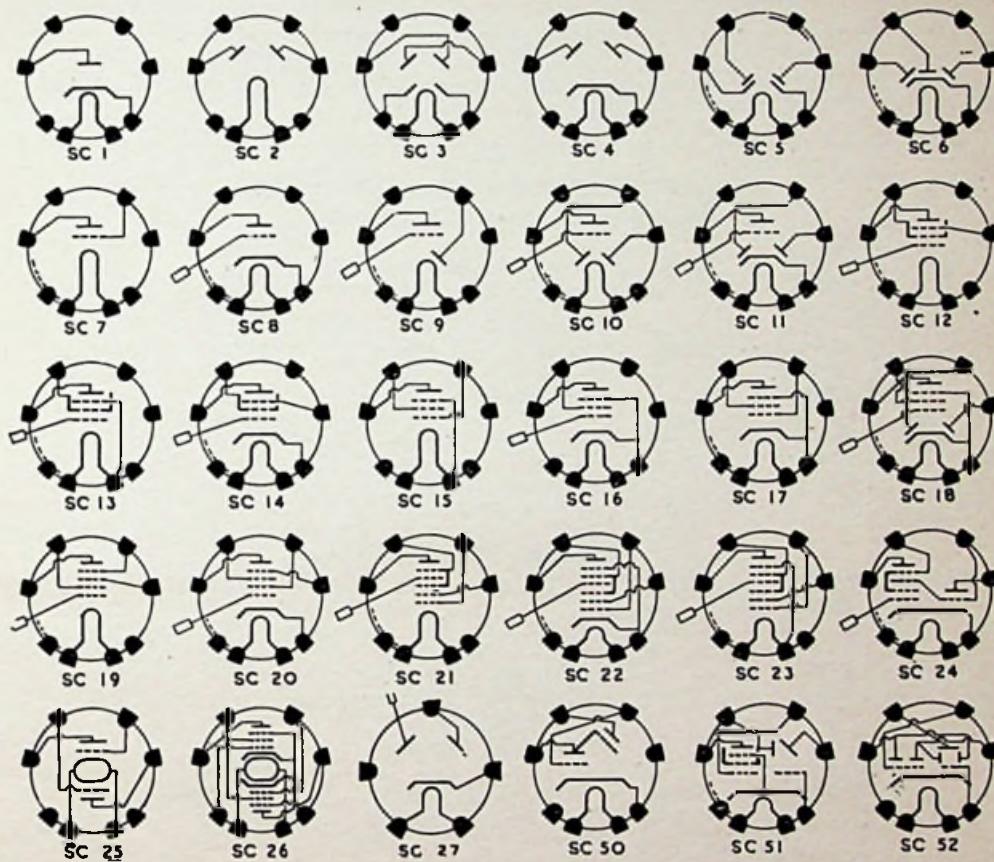


Note.—All connections are as viewed from underside of valve base.



Note.—All connections are as viewed from underside of valve base.

SIDE CONTACT



Note.—All connections are as viewed from underside of valve base.

VALVE EQUIVALENTS GUIDE

The following suggested equivalents are given in good faith, and it is realised that some possible equivalents may be missing and that errors may exist, since sources of information may on occasion be inaccurate. Certain valves listed may not be obtainable, but the information will nevertheless be of value to those who possess such types. Valves enclosed by a bracket are mutually interchangeable.

Equivalents of Continental valves are not included; for such information and for equivalents of British, U.S.A., German and Italian war-time service valves, it is suggested that reference be made to the *Radio Tube Vade-Mecum*, by P. H. Brans, published at 28 Rue du Prince Leopold, Anvers, Belgium. This book has all data in English, and may be purchased in this country.

<i>Brimar</i>	<i>Cossor</i>	<i>Ever Ready</i>	<i>Ferranti</i>	<i>Marconi Osram</i>	<i>Mazda</i>	<i>Mullard</i>
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2-V BATTERY (BRITISH AND U.X. BASES)

Screened Grids and H.F. Pentodes

32E	32E	—	—	—	—	32
5B1	215SG	K40B	—	S23	S215A	PM12
—	220SG	—	—	S22	S215B	PM12A
—	—	—	—	S24	SG215	—
—	—	—	—	S21	—	—
—	220VS	K40N	VS2	VS2	S215VM	PM12M
—	220VSG	—	—	VS24	—	—
—	210SPT	—	—	Z21	SP210	SP2
—	210VPA	K50M	VPT2	VP21	VP210	VP2
—	210VPT	—	—	W21	VP215	—
1A4E	1A4E	—	—	—	—	1A4
34E	34E	—	—	—	—	34
15E	15	—	—	—	—	15

Frequency Changers

—	210PG	K80A	VHT2A	X21	—	FC2
—	210SPG	—	—	X22	—	FC2A
—	210PGA	K80B	VH2	—	—	FC2A
—	—	—	—	X24	—	TH2
1A6	1A6E	—	—	—	—	1A6
1C6	1C6E	—	—	—	—	1C6

Diodes and Triodes

—	220DD	—	H2D	—	DD207	2D2
—	210DDT	K23B	—	HD22	HL21DD	TDD2A
—	—	K23A	—	HD24	L2DD	TDD2
—	—	—	—	HD23	L21DD	—
2102	2102	—	—	—	—	2102

Triodes

HLB1	210RC	K30A	—	H2	H2	PM1A
—	210HL	K30K	—	HL2	HL2	PM2HL
—	210HF	K30C	—	HL210	HL210	PM1HF
(1) {	210LF	K30D	L2	L210	L2	PM1LF
PB1	210DET	K30B	—	L21	L210	PM2DL
—	—	K30E	—	L2/B	—	PM2DX
30	215P	—	—	P215	P215	PM2
—	220PA	K30G	L2	LP2	P220	PM2A
—	230XP	—	—	P2	P220A	PM202
30	30	—	—	—	—	30

(1) Certain valves in this group have special anti-microphonic properties. These are not necessarily shared by all the equivalents shown.

Brimar	Cossor	Ever Ready	Ferranti	Marconi Osram	Mazda	Mullard
Output Pentodes and Tetrodes						
PENB1	220HPT	K70B	—	PT2 KT2	PEN220	PM22A
—	220OT	—	—	—	PEN220A	PM22
—	220PT	—	—	—	—	PM22C
—	—	K70D	—	KT21 KT24	PEN231	PM22D
2101	2101	—	—	—	—	2101
Class B Output						
—	220B	K33A	HP2	—	PD220	PM2B
—	240B	K33B	—	—	PD220A	PM2BA
19	19	—	—	—	—	19
Q.P.P.						
—	240QP	K77A	QPT2	QP21	QP23	QP22B
—	—	—	—	—	QP240	QP22A
2103	2103	—	—	—	—	2103
1·4-V BATTERY (OCTAL BASES)						
H.F. Pentodes						
IN5G	1N5G	1N5G	—	Z14	SP141	1N5G DF33
Frequency Changers						
1A7EG	1A7G	1A7G	—	X14	FC141	DK32 1A7G
Diode Triode						
1H5G	1H5G	1H5G	—	HD14	H141D	DAC32 1H5G
Output Pentodes						
1C5G	1C5G	1C5G	—	N14	PEN141	DL35 1C5G
4-V A.C. MAINS (BRITISH BASES)						
Screened Grids						
SGA1	MSG/HA	—	—	MS4	AC/S2	S4VA
—	41MSG	—	—	—	—	S4V
—	MSG/LA	—	—	MS4B	AC/SG	S4VB
—	MVSG	A40M	—	VMS4	AC/S1VM	MM4V
—	—	—	—	VMS4B	AC/SGVM	VM4V
H.F. Pentodes						
8A1	MS/PEN	A50A	SPT4	MSP4	AC/S2PEN	VP4
—	MS/PENA	—	SPT4A	—	—	—
9A1	MVS/PEN	A50M	VPT4	VMP4	AC/SP3	TSP4
—	—	A50N	VPT4A	VMP4G	AC/VP1	VP4
—	MVS/PENB	A50P	VPT4B	—	—	VP4A
—	MS/PENB	A50B	—	W42	AC/VP2	VP4B
—	—	—	—	—	—	SP4B
Frequency Changers						
20A1	41STH	A36A	—	X41	AC/TH1	TH4
—	—	A36B	—	—	—	—
15A2	41MPG	A80A	VHT4	X42	—	FC4
—	—	—	—	MX40	—	—
—	4THA	A36C	—	—	—	TH4B

Brimar	Cossor	Ever Ready	Ferranti	Marconi Osram	Mazda	Mullard
Diodes and Diode Triodes						
—	DD4	A20B	—	D41	AC/DD V914	2D4A
11A2	DDT	A23A	H4D	MHD4 DH42	AC/HLDD	TDD4
Triodes						
HLA1	41MH	A30B	—	MH41	AC2/HL	904V
HLA2	41MHF	A30D	D4	MH4	—	354V
—	41MHL	—	—	—	—	244V
—	41MLF	—	—	MHL4	—	164V
—	41MP	—	—	ML4	AC/P	104V
PA1	41MXP	—	—	—	AC/P1	TT4
—	—	—	—	—	PA40*	O54V
—	—	—	—	DA30*	DO26*	DO26*
—	4XP*	S30C*	LP4*	PX25A*	V503*	DO30*
—	—	—	—	PX4*	PP3/250*	ACO44*
—	2P†	—	—	PX25*	PP5/400*	DO24*
—	2XP†	—	—	—	PA20†	ACO42†
Output Pentodes and Tetrodes						
7A2	MP/PEN	A70B	—	MPT4	AC/PEN	PEN4V
—	—	—	—	MKT4	—	PEN4VA
7A3	42MP/PEN	A70D	PT4	KT42	—	PEN4VB
—	42OT	A70C	—	KT41	AC2/PEN	PEN4A4
—	42OTDD	—	PT4D	N41	—	—
—	—	A27D	—	DN41	AC2/PENDD	PEN4DD
PENAL	PT41	—	—	PT4	—	PM24
—	PT41B	—	—	PT25	—	PM24B
—	—	—	—	PT25H	—	PM24E
A.C./D.C. MAINS (BRITISH BASES)						
Screened Grids and H.F. Pentodes						
—	—	—	SPTS	—	SP1320	—
—	—	—	VPTA	—	VP1320	—
—	—	—	—	—	VP1321	—
9D2	13VPA	C50N	—	—	VP1322	VP13C
Frequency Changers						
—	302THA	C36N	—	—	—	TH30C
—	—	—	—	—	TH2321	TH22C
—	—	—	—	X31	—	TH13C
15D1	202STH	C36A	—	—	—	TH21C
—	13PGA	C80B	—	—	—	FC13C
Diodes and Triodes						
10D1	—	C20C	ZD	—	DD620	2D13C
11D3	13DHA	C23B	HAD	—	HLDD1320	TDD13C
Triodes						
4D1	—	C30B	DA	—	HL1320	HL13C
7D3	40PPA	—	—	—	—	—
Output Pentodes						
7D8	—	—	PTZ	—	PEN1340	PEN13C
—	402PEN	—	—	—	—	—
—	402OT	—	—	—	—	—
7D6	—	C70D	—	—	PEN3520	PEN36C
—	—	—	—	—	PEN3820	—

* Directly heated. Bias resistors in use should be checked for suitability.
 † 2-volt filaments. Same remarks as * apply.

<i>Brimar</i>	<i>Ferranti</i>	<i>Cossor</i>	<i>Marconi Osram</i>	<i>Mullard</i>	<i>U.S.A.*</i>
INTERNATIONAL MAINS VALVES (U.X. AND OCTAL BASES)					
36	—	36E	—	36	36
36E	—	—	—	—	—
24A	—	24E	—	24A	24A
24E	—	—	—	—	—
Screened Grids †					
78	—	78E	—	78	78
78E	—	—	—	—	—
6U7G	—	—	KTW63	6K7G	6U7
39/44	—	39/44E	—	39/44	39/44
39/44E	—	—	—	—	—
6K7G	6K7G	6K7G	—	6K7G	6K7
6K7EG	—	—	—	—	—
77	—	77E	—	77	77
77E	—	—	—	—	—
6B8G	—	6B8EG	—	6B8G	6B8
6B8EG	—	—	—	—	—
6B7	—	6B7E	—	6B7	6B7
6B7E	—	—	—	—	—
6J7G	—	—	Z63	—	6J7
Frequency Changers					
6A7E	—	6A7E	—	6A7	6A7
6A8G	6A8G	6A8EG	X63	6A8G	6A8
6A8EG	—	—	—	—	—
6L7G	—	—	X64	—	6L7
—	—	—	X65	—	6K8
Diodes, Diode Triodes and Diode L.F. Pentodes					
6H6G	—	6H6G	D63	6H6G	6H6
6R7G	—	6R7G	DL63	6R7G	6R7
75	—	75	—	75	75
85	—	85	—	85	85
6Q7G	6Q7G	6Q7G	DH63	6Q7G	6Q7
6B8G	—	6B8EG	—	6B8G	6B8
6B8EG	—	—	—	—	—
6B7	—	6B7E	—	6B7	6B7
6B7E	—	—	—	—	—
Triodes					
6K5G	—	6K5G	—	6K5G	6K5
6J5G	—	—	L63	6J5G	6J5
6F5G	—	—	H63	—	6F5
6B4G	—	6B4G	—	—	6B4
6A3	—	6A3	—	6A3	6A3
76	—	76	—	76	76
27	—	27	—	27	27
37	—	37	—	37	37
45	—	45	—	45	45
Output Pentodes and Tetrodes †					
6L6G	—	—	KT66	EL35	6L6
6F6G	6F6G	6F6EG	KT63	6F6G	6F6
6F6EG	—	—	—	—	—
41E	—	41E	—	41	41
42	—	42E	—	42	42
42E	—	—	—	—	—
18	—	18E	—	18	18
43E	—	43E	—	43	43
47	—	47E	—	47	47
47E	—	—	—	—	—
—	—	—	KT32	25L6G	25L6

* In this column only the significant characters are given; the suffix characters denote type of envelope.

† Valves with an E suffix are not necessarily equivalents when used as oscillators.

Brimar	Ferranti	Cossor	Marconi Osram	Mullard	U.S.A.*
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Class B Output					
6N7G 79	—	6N7G 79	—	6N7G 79	6N7 79

MISCELLANEOUS

Cossor	Marconi Osram	Ferranti	Marconi Osram	Cossor	Mullard
DVSG	VDS	VHTS	X30	OM4	EBC33
DHL	DH	VPTS	W31	OM5	EF36
DDT16	DHD	DS	H30	OM6	EF39
DP/PEN	DPT	HSD	DH30	OM9	EL32
—	—	PTA	KT30	OM10	ECH33

Brimar	Cossor	Ever Ready	Ferranti	Marconi Osram	Mazda	Mullard
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RECTIFIERS

A.C. (British Bases)

R2	—	506BU 43IU	S11A A11B A11D	— — R4	U10 — U12	UU2 UU4 —	DW2 IW3 IW4/350
R3	—	44IU	A11C	—	MU12 MU14	UU5 —	DW3 DW4/35
—	—	442BU	S11D	—	U14	UU120/350 UU120/500	—
—	—	460BU	—	—	GU1 GU50	UU30/250	DW4 DW4/500
—	—	408BU	—	—	U18	—	RG1/250 RG1/240
—	—	4/100BU	—	—	—	—	FW4/500
—	—	415U	—	—	—	—	—

A.C./D.C. British Bases

ID5	—	40SUA	—	C10B	RZ	—	U4020	—	UR1C
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U.X. AND OCTAL BASES

5X4G	5X4G	—	—	—	U52	—	—	5U4G
—	—	—	—	—	U50	—	—	5X4G
5Y4G	5Y4G	—	—	—	—	—	—	5Y3G
5Z3	5Z3	—	—	—	—	—	—	5Y4G
6ZY5G	6ZY5G	—	—	—	—	—	—	5Z3
25RE	25RE	—	—	—	—	—	—	—
35RE	35RE	—	—	—	—	—	—	—
25Z5	25Z5	—	—	—	—	—	—	25Z5
—	80	—	—	—	—	—	—	80
84	84	—	—	—	—	—	—	84

* In this column only the significant characters are given; the suffix characters denote type of envelope.

