

STAGE LIGHTING CONTROLLER
Cellular Logic a Analogue To Digital Visual Hi-Fi * Magnetic Amplifiers

## CHROMATHEQUE 5000

 5 CHANNEL LIGHTING EFFECTS SYSTEM

All kits also available as separate packs (e g P C B, component sets, hardware sets, etc) Prices in FREE CATALOGUE

COMPLETE KIT ONLY
$\mathbf{£ 4 9 . 5 0}$ + VAT!

This versatile system featured as a constructional article in ELECCTRONICS TODAY INTERNATIONAL has 5 frequency channels with individual level controls on each channel. Control of the lights is comprehensive to say the least You can run the unit as a straightforward sound-to-light or have it strobe all the lights at a speed dependent upon music level or front panel control or use the internal digital circuitry which produces some superb random and sequencing effects. Each channel handles up to 500 W and as the kit is a single board design wiring is minimal and construction very straightforward
Kit includes fully finished metalwork, fibreglass PCB, controls, wire, etc. - Complete right down to the last nut and bolt!

## LAST MONTH'S FRONT COVER FEATURE! 100 WATT (rms into 8 $\Omega$ ) MIXER / AMPLIFIER



COMPLETE KIT ONLY $£ 49.90$ + VAT

Parts to build power amp module only. (PCB, res, caps, $\mathrm{s} / \mathrm{cs}$ ) $\quad \mathbf{£ 1 0 . 6 0}+$ VAT Kit includes fully finished metalwork,
Custom designed toroidal transCustom designed toroidal transformer with mounting clamp $\mathbf{£ 1 0 . 5 0}+$ VAT Parts for power supply only (caps, rects, fuses, F. hoiders)

## TRANSCENDENT 2000 SINGLE BOARD SYNTHESIZER

LIVE PERFORMANCE SYNTHESIZER DESIGNED BY CONSULTANT TIM ORR (FORMERLY SYNTHESIZER DESIGNER FOR EMS LIMITED) AND FEATURED AS A CONSTRUCTIONAL ARTICLE IN ELECTRONICS TODAY INTER NATIONAL.
 modulation, a VCF with both low and high pass outputs and a separate dynamic sweep control, a noise generator and an ADSR envelope shaper There is also a slow oscillator, a new pitch detector. ADSR repeat, sample and hold, and special circuitry with precision components to ensure tuning stability amongst its many features.

The kit includes fully finished metalwork, fully assembled solid zeak cabinet. filter sweep pedal. professional quality components (all resistors either $2 \%$ maral oxide or $1 / 2 \%$ metal triml) and it really is complete - right down to the last nut and bolt and
last piect of wirel There is even a 13 a plug in the kit - you nead buy absolutely no more parts before plugging in and making great music) Vou nead buy absolutely no are on the one professional quatity fibreglass PCB printed with component locations All the controls mount directly on the main board, all connections to the board are made with connector plugs and contruction is so simple it can be built easily in a few evenings by almost anyone capable of neat solderingl when finished you will seiling for befween $£ 500$ and $£ 7001$

## COMPLETE KIT ONLY $£ 172.00$ + VAT!

Comprehensive handbook supplied with all complete kits! This fully describes construction and tells you how to set up your synthesizer with nothing more elaborate than a multi-meter and a pair of ears!


Cabinct size $24.6^{\prime \prime} \times 15$. $^{\prime} \times 4.8^{\prime \prime}$ (rear) 3.4" (front)
ORDERING INFORMATION AND MORE KITS ON PAGE 6


Strike a Light p27


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## PROJECTS

## INFORMATION

All the latest
Modern conversion techniques
Cellular Logic Image processors Ok?
Magnetic Amplifiers, watts that?
News from MPU's
Visual Hi-Fi but how does it sound?
Your circuits, not bad either
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OUR THANKS TO LAMBETH TOWN HALL FOR THE COVER PICTURE THIS MONTH, A SCENE FROM THEIR PRODUCTION OF TREASURE ISLAND.


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## DASS WOMOMONTEG

## 74 SERIES TTL IC's

| Type | Price | Type | Price | Type | Price | Type | Price | Type | Price |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7400 | ¢0.07 | 7427 | E0.21 | 7472 | ع0.19 | 74107 | c0. 22 | 74165 | £0.65 |
| 7401 | c0.09 | 7428 | E0.25 | 7473 | ¢0.22 | 74110 | E0.35 | 74166 | ¢0.75 |
| 7402 | E0.09 | 7430 | E0.08 | 7474 | E0. 22 | 7411 | ¢0.55 | 74157 | £2.00 |
| 7403 | E0.09 | 7432 | c0. 20 | 7475 | ¢0.27 | 74118 | ¢0.75 | 74174 | ¢0.60 |
| 7464 | c0. 09 | 7433 | ${ }^{\text {E }}$. 28 | 7476 | £0. 22 | 74119 | E1.10 | 74175 | ¢0. 60 |
| 7405 | E0.09 | 7437 | ¢0.20 | 7480 | c0.40 | 74121 | £0.22 | 74176 | ¢0.55 |
| 7406 | ¢0. 22 | 7438 | ${ }^{5} 0.20$ | 7481 | ¢0.80 | 74122 | E0.35 | 74177 | ¢0.55 |
| 7407 | E0.22 | 7440 | E0.18 | 7482 | ¢0.65 | 74123 | ¢0.38 | 74180 | £0. 80 |
| 7408 | E0.12 | 7441 | ${ }_{60.45}$ | 7483 | E0. 55 | 74136 | c0.50 | 74181 | £1.25 |
| 7409 | ع0.12 | 7442 | ${ }^{10.38}$ | 7484 | ¢0.82 | 74141 | ¢0.50 | 74182 | ${ }^{\text {co. }}$. 55 |
| 7410 | E0. 09 | 7443 | c0.68 | 7485 | ¢0.65 | 74145 | ¢0.54 | 74184 | £1.00 |
| 7411 | ¢0.15 | 7444 | ¢0.68 | 7486 | £0.22 | 74150 | E0.65 | 74190 | ¢0.68 |
| 7412 | En 14 | 7445 | ¢0.64 | 7489 | ¢1.60 | 74151 | ¢0.45 | 74191 | E0.68 |
| 7413 | 50.22 | 7446 | ¢0.60 | 7490 | ع0,30 | 74153 | £0.45 | 14192 | £0.65 |
| 7414 | c. 0.45 | 7447 | E0.45 | 7491 | ع0.60 | 74154 | ¢0.80 | 74193 | £0.80 |
| 7416 | ¢0.22 | 7448 | E0. 52 | 7492 | ¢0.32 | 74155 | E0.48 | 14194 | ¢0.55 |
| 7417 | ${ }_{50} 0.22$ | 7450 | ¢0.09 | 7493 | [0. 28 | 74155 | ¢0.48 | 74195 | E0.53 |
| 7420 | ¢0.09 | 7451 | E0.09 | 7494 | ع0.70 | 74157 | ${ }_{\text {c }} 0.48$ | 74196 | ¢0.60 |
| 7421 | ¢0, 19 | 7452 | c0.09 | 7495 | ${ }^{\text {c0. }}$ ( 45 | 74160 | £0.55 | 74197 | ¢0.58 |
| 7422 | ¢0.15 | 7453 | ¢0.09 | 7496 | ${ }^{\text {c0. }}$. 48 | 74161 | ${ }^{\text {c }} 0.60$ | 74198 | £1.00 |
| 7423 | ¢0.20 | 7454 | E0.09 | 74100 | E0.80 | 74162 | ${ }^{1} 0.60$ | 74199 | ¢1.00 |
| 7425 | c.0.18 | 7460 | ¢0.09 | 74104 | ${ }^{\text {co, }} \mathbf{3 5}$ | ${ }_{7} 74163$ | ${ }_{\text {¢ } 0.60}$ | 74279 | E1.00 |
| 7426 | ع0.21 | 7470 | c0. 24 | 74105 | ¢0.35 | 74164 | £0.65 |  |  |

CMOS IC's

| Type | Price | Type | Price | Type | Price | Type | Price |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CD 4000 | E0.12 | CD4017 | $\underline{60.65}$ | CD403 ${ }^{\text {c }}$ | £1.60 | CD4055 | E1.00 |
| CD400 | E0.13 | CD4018 | ¢0.70 | CD4035 | co. 90 | CC4056 | E1.15 |
| CD4002 | E0.13 | CO4019 | ¢0.35 | C04037 | ¢0.78 | CD4069 | ${ }^{50.15}$ |
| CD4006 | E0.80 | CD4020 | ¢0.80 | CD4040 | ¢0.78 | CD4070 | co. 15 |
| CD4007 | E0. 14 | CO4021 | ¢0.75 | CD4041 | ¢0.68 | CD 4071 | c0.15 |
| CD4008 | ¢0.80 | CD4022 | ¢0.75 | CD4042 | £0.68 | CD4072 | c0. 15 |
| CD4009 | c0.40 | CO4023 | £0.13 | CD4043 | ¢0.78 | CD4081 | ¢0.15 |
| CD4010 | co. 42 | CD4024 | ¢0.55 | CD4044 | c0.78 | CD4082 | ¢0.16 |
| CD4011 | ¢0.13 | CD4025 | ¢0.13 | CD4045 | ¢1.15 | CD4510 | c0. 80 |
| CD4012 | E0.14 | CD4026 | ¢1.00 | CD4046 | ¢0.95 | CD4511 | c0.80 |
| CO4013 | ¢0.35 | CD4027 | ¢0.45 | CD4047 | ¢0.75 | C04516 | ¢0.85 |
| CD4014 | ¢0.70 | CO4028 | ¢0.60 | CD4049 | ${ }^{60.35}$ | CD4518 | c0.85 |
| CD4015 | £0.70 | CD4029 | £0.40 | CDA050 | ¢0.35 | CD4520 | ¢0.85 |
| CD4013 | £0.35 | CD4030 | £0.55 | CD4054 | £0.95 |  |  |

SPECIAL OFFER！ COMPONENT PAKS

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|  | －issmed elactrolytics |  |
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| O－ | E－ | 400 |
| 为 | －5 Assorted fuses 100 ma 5 amp | ${ }_{40 \mathrm{p}}^{40}$ |
|  | $50 \% \mathrm{~W}$ resistors mixed val | 40 p |
|  | 30 meres stranded wire |  |
| ミ＇ | 120 \％walt resistors |  |
|  | 1978 Pros．Our mix | 60p． |
| ミ｀＊ | $1201 /$ watt resistors，Pre－formed |  |
| ミ－2 | 978 Prod．Mixed vailues | 60p． |
|  | Stis wart resis |  |
| s．c3 | Range fo enms． 10 |  |
|  | Renge 100 ohms 10 | E2．00 |
| s．cs | $60 \mathrm{Low} \mathrm{ohms} 1 / 8$ watt resistors |  |
| $\bigcirc \cdot 55$ | 10.100 ohms | 60p＊ |
|  | cow ohms $/ 2 \mathrm{w}$ |  |
| $\begin{aligned} & 5: 06 \\ & 5: 07 \end{aligned}$ | Mixed wire |  |
|  | ${ }_{20} 5$ Tontan wium bend cand resistors |  |
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| 5：28 | High－quality electroivics 10 |  |
|  | volage range 1550 V ． |  |
| 18204 | 880 mak Contain | ．00 |
|  | toil caps | E1．00＇ |

## POTENTIOMETERS

## Slider $\mathbf{4 0 m m}$ TRAVEL



S38 Mixed slider pots－various values
and sizes；our mix


## WIREWOUND

590 Wirewound Pots Linear 1 Wart
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2.5 Kkir
each 60p＊
6 mm Shaft
$5924 \times 100 \times \mathrm{Lin}$ ．
$5934 \times 100 \mathrm{KLog}$
1618625 Pre－sets Assorted Vilue


## ZENER PAKS

No S55 $\quad \begin{gathered}20 \text { mixed values } 400 \text { niW Zener } \\ \text { diodes } 3 \text { lov }\end{gathered}$
No． 556 mixed values 400
No． $557 \quad \begin{gathered}\text { diodes } 11-33 \mathrm{~V} \\ 10 \text { mixed values } 1 \mathrm{~W} \text { Zener }\end{gathered}$
No． 558 diodes $\mathbf{3}$－10V
No． 558

$$
\begin{aligned}
& 10 \text { mixed values } 1 \mathrm{~W} \text { Zener } \\
& \text { diodes } 11-33 \mathrm{~V}
\end{aligned}
$$

SILICON POWER TRANS．N．P．N



| Crystal Ear Pieces <br> S126 Less plug | £0．20 |
| :---: | :---: |
| Plugs for above |  |
| No 16106.25 plastc No 169735 plastic | $\begin{gathered} £ 0.09 \\ £ 0.11 \end{gathered}$ |
| Mono Crystal Cartridge |  |
| Nickel Cadnium Rechargeable Batteries，$1.25 \mathrm{v}$ |  |
| S128． 35000 Cell size $=$ U2 | ¢2．50 |
| S129． 900 C Cell size $=1 / 2011$ | ¢0．90 |
| build nickel cadnumen charger | £3．50 |


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## TRANSISTORS

| Type | Price | Type | Price | Type | Price | Type | Price | Type | Price |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AC107 | 25p | BC177 | 12p | BF194 | 9 p | TIP32A | 34p | 2N1613 | 15p |
| AC126 | 14p | BC178 | 12p | BFi95 | －9p | TIP32B | 35p | 2N1711 | 15p |
| AC127 | 16p | BC179 | 12p | BF196 | －12p | TIP32C | 36p | 2N1B93 | 28p |
| AC128 | 16 p | BC182 | ${ }^{9 p}$ | BF 197 | ＊12p | TIP41A | 34p | 2N2218 | 15p |
| AC128K | 24p | BC1B2L | 9p | BF200 | 25p | TIP41B | 35 p | 2N2218A | 18p |
| AC176 | $16 p$ | BC183 | ＊9p | BFX29 | 22p | TIP41C | 36p | 2N2219 | 15p |
| AC176K | 24p | BC1B3L | －9p | BFX84 | 18p | TIP42A | 36p | 2N2219A | 18p |
| AC187 | 16p | BC184 | ＂9p | BFY50 | 12p | TIP42B | 37p | 2N2221 | 15p |
| AC187K | 26p | BC184L | －9p | BFY51 | 12 p | TIP42C | 38p | 2N2221A | 16p |
| AC188 | $16 p$ | BC212 | $\cdot 10 \mathrm{p}$ | BFY52 | 12p | TIP2955 | 65p | 2N2222 | 15p |
| AC188K | 26p | BC212L | ${ }^{10} 10$ | BFF2 | 12p | TIP 3055 | 42p | 2N2222A | 16p |
| AD161／ |  | BC213 | $\cdot 10 p$ | MPSA05 | －22p | 2TX107 | ＊ 6 p | 2N2369 | 10p |
| 162 MP | 80p | BC213L | －10p | MPSA06 | ＂22p | 2TX108 | \％p | 2N2904 | 14p |
| AF139 | 30p | BC214 | ＊10p | MPSA55 | ＊22p | ZTX109 | ＊7p | 2N2904A | 15p |
| AF239 | 30p | BC214L | ＊10p | MPSA56 | ＊22p | $21 \times 300$ | ＊7p | 2N2905 | 14p |
| BC107 | 6 p | BC251 | －10p |  |  | $21 \times 301$ | ${ }^{7} \mathbf{7 p}$ | －2N2905A | 15p |
| BC108 | 6p | BCY70 | 12p | $0 \mathrm{OC45}$ | 12p | $21 \times 302$ | ＂9p | 2N2906 | 12p |
| BC109 | 6 p | BCY7 | 12p | $0 \mathrm{OC71}$ | 12p | ZTX500 | ＂8p | 2N2906A | 14p |
| BC1 18 | $\cdot 10 p$ | BCY72 | 12p | OC72 | 9p | 2TX501 | $\cdot 10 p$ | 2N2907 | 12p |
| BC147 | ＂8p | BD 115 | 40p | 0 O 75 | 12p | 2TX502 | $\cdot 12 p$ | 2N2907A | 13p |
| BC148 | ＊8p | BD131 | 35p | OC81 | 10p | 2N696 | $\cdot 10 \mathrm{p}$ | 2N2926G | 8p |
| BC149 | －8p | BD132 | ＊37p | 0 ¢8 | 14p | 2N697 | 10p | 2N2926Y | ＇7p |
| BC1 54 | ＊16p | BF115 | 17p | TIP29A | 35p | 2N706 | 7 p | 2N3053 | 12p |
| BC157 | ＂9p | BF167 | 19p | TIP29B | 36p | 2N706A | 8 p | 2N3055 | 35p |
| BC158 | ＊9p | BF173 | 20p | T／P29C | 38p | 2N70B | 8 p | 2N3702 | 7p |
| BC159 | ＂9p | BF1B0 | 25p | TIP30A | 36p | 2N1302 | 12p | 2N3703 | 7 p |
| BC169 | ＊10p | BF181 | 25p | TIP30B | 37p | 2N1303 | 15p | 2N3704 | ＊ 6 p |
| BC170 | 6 p | BF 182 | 25p | TIP30C | 38p | 2N1304 | 15p | 2N3903 | 11p |
| 8 BC 171 | ＊ 6 p | BF183 | 25p | TIP31A | 32p | 2N1307 | 18p | 2N3904 | 11p |
| 8C172 | ${ }^{6} \mathbf{p}$ | BF184 | 25p | TIP318 | 33p | 2N1308 | 22p | 2N3905 | 11p |
| BC173 | 7p | BF185 | 25p | TIP31C | 34p | 2N1309 | 22p | 2N3906 | －11p |

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## As featured in Electronics Today International

 400W rms continuous - 800W peak! $0.03 \%$ THD at FULL power! PLUS all the following features too!* Each channel totally independent with its own stabilised power supply driven by custom designed TOROIDAL transformers!
* Inherent reliability - monster heat sinks for cool running at the hottest venues -- electronic open and short circuit protection!
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* Professional quality components, sturdy $19^{*}$ rack mounting chassis complete with sleeve and teet for free standing work too
* Easy to build - plenty of working space with ready access to all components, minimal wiring. extensive instruction suitable for both experience constructors and newcomers to electronics.
$\star$ Value for money - quality and performance comparable with ready-built amplifiers costing over £6001


## DE LUXE EASY TO BUILD LINSLEY HOOD 75W STEREO AMPLIFIER £99.30 + VAT

This easy to build version of our world-wide acclaimed 75 W amplifier kit based upon circuit boards interconnacted with gold plated contacts resulting in minimal wiring and construction delightfully straightforward. The design was published in Hi-Fi News and Record Review and features include rumble filter, variable scratch filter, versatile tone controls and tape monitoring whilst distortion is less than $0.01 \%$


## WIRELESS WORLD FM TUNER £70.20 + VAT

A pre-aligned front end module makes this Wireless World published design very simple to construct and adjust without special instruments. Features include an excellent a.m. rejection. push-button station selection as well as infinitely variable tuning and a phase locked loop stereo decoder incorporating active filters for "birdy" suppression.

LINSLEY-HOOD CASSETTE DECK £79.60 + VAT
This design, published in Wireless World, although straightforward and relatively low cost provides a very high standard of performance. There are separate record and replay amplifiers and switchable equalisation together with a choice of bias levels are also provided. The mechanism is the Goldring-Lenco CRV with electronic speed control.

T20 + 20 20W STEREO AMPLIFIER £33.10 + VAT
This kit, based upon a design published in Practical Wireless, uses a single printed circuit quality amplifiers. A 30 watt version of this kit $(T 30+30)$ is also available for $\mathbf{£ 3 8 . 4 0}+$ VAT

## POWERTRAN SFMT TUNER £35.90 + VAT

Thes is a simple low cost design which can be constructed easily without special alignment equipment but which still gives a first-class output suitable for feeding any of our very popular amplifiers or any other high quality audio equipment. A phase-locked-loop is used for stereo decoding and controls include switchable afc, switchable muting and push-button channel
selection (adjustable by controls on the front panel). This unit matches well with the $T 20+20$ selection (adjustable by controls on the front panel). This unit matches well with the $\mathrm{T} 20+20$ and $\mathrm{T} 30+30$ amplifiers.

## WWII TUNER $£ 47.70$ + VAT

This cost reduced model of our highly successful Wireless World FM Tuner kit was designed to complement the T20 +20 and T30 +30 amplifiers and the cabinet size. front panel format and electrical characteristics make this tuner compatible with either Facilities included are pre-aligned front-end module, switchable afc, adjustable switchable muting, LED tuning indication and both continuous and push-button channel selection (adjustable by controls on the front panel)


COMPLETE KITS: Our complete kits really are complete. All of the projects shown on this pagerare supplied with fully finished metalwork. ready assembled high quality teak veneer cabinet. cables, nuts, bolts, etc. and full instructions - in fact everything!

All of the kits shown on this page are available as separate packs (except the Powertran SFMT Tuner) for those customers who wish to spread their purchase or perhaps make the. cwr cabinets or metatwork. Prices are given in our FREE CATALOGUE.
PRICE STABILITY. Order with confidence irrespective of any price changes. We will honour all prices in this advertisement until April 30th, 197 g , if the March 1979. issue is mentioned with your order. Errors and VAT rate changes excluded. EXPORT ORDERS: No VAT Postage charged at actual cost plus 50 p handling and documentation
U.K. ORDERS.Subject to $12 \frac{1}{2} \%$ surcharge for VAT ${ }^{*}$ (i.e. add $1 / 8$ to the price). No charge is made for carriage. or at current rate if changed
SECURICOR DELIVERY: For this, optional service (U.K. mainland only) add 2. 50 (VAI inclusive) per kit

SALES COUNTER: If you prefer to collect your kit from the factory, call at Sales Counter (at rear of factory) Open $9 \mathrm{a} . \mathrm{m} .-4.30$ p.m. Monday-Thursday.

# POWERTRAN ELECTRONICS 

[^0]
# news digest....... <br> \section*{HAND-HELD GAMES} 



## VIDEO DISC REVISITED

It looks like the video disc is rearing its domestic head again. Philips have introduced a 60
minutes per side system, release and price for UK market have yet to be announced, but be prepared for yet another compatability war as other manufacturers join the fray.


## SINCLAIR ANNOUNCE UK MICROVISION

A 'UK only standard' version of the top selling Microvision has been developed by Sinclair Radionics. Outwardly it has the same dimensions as the International version, but fewer controls. Good news for bank managers too, it costs less than half the previous version at less than $£ 100$. Further details from Sinclair Radionics, London Road, St Ives, Huntingdon, Cambs. PE17 4HJ.

## STRIKE A BLOW

We are reliably informed that due to the industrial action within the transport industry, copies of ETI have not reached some areas for last month's (February) issue. We apologise to our readers for this and hop you'll bear with us through the trouble.

Owing to the continuing - as we go to press troubles, this issue too may be delayed. In some cases this may well be severe. If you read this later than you would normally have done so - thank you for sticking it out, and we promise normal service will be resumed as soon as possible.

> Ron Harris
> Editor


After the inroads made into the leisure market by TV video games, a new generation of hand-held calculator-style games seem to be making their way across the Atlantic. The 'AMAZE-A-TRON' (groan) is a micro-based game specifically aimed at the 5 years and up age range. It is basically a maze game
with a claimed one million variations. Also in the pipeline are 'ZAP' a missile game, 'DIGITS' a code game, and 'LIL GENIUS', a teaching type calculator. Prices will range from $£ 9-£ 18$ and will be marketed by Spectrum Electronic Games, 113-115 Gloucester Road, London SW7 4TE.

## HOT STUFF



Details of a new pocket-sized thermometer, 'computerised' no less, have just arrived. Designated the ITS there are four models in the range, two :over the range $0-110^{\circ} \mathrm{C}$ and the other two from $-35^{\circ} \mathrm{C}$ to $149^{\circ} \mathrm{C}$. The LED display can handle Fahrenheit as well as Centigrade. More
than 25 interchangeable probes are available for various applications, and its rugged high impact aluminium case is ideal for field use. Contact British Rototherm Co Ltd for further details at Kenfig Industrial Estate, Margam, Port Talbot, West Glam. SAl3 2PW (South Wales).

## WHIITOMD EIETITVIUS $33 / 35$ CARDIFF ROAD, WATFORD, HERTS, EN MAIL ORDER, CALLERS WELCOME.



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SILVER MiCA Nalues in pF$) 3$ 3-3. 4.7
$6.8,10,12,18,22.33,47,50,68,75$
$82.85,100.130 .150 .2$
$250.300,330,360.390$
$\begin{array}{ll}1000,1800,2000,2200 & \begin{array}{l}16 p \text { each } \\ 20 p \text { each }\end{array}\end{array}$
POLYSTYRENECAPACITORS:
10pF to inf Bp ; $1.5 \cap$ to 47 nF 10p
MINIAYURE TYPE TRIMMERS
$2.5-6 \mathrm{pF}: 3.10 \mathrm{pF} ; 10-40 \mathrm{pF}$
$5.25 \mathrm{pF}: 5.45 \mathrm{pF}: 60 \mathrm{pF}, 88 \mathrm{pF}$.
COMPRESSION TRIMME
$3-40 \mathrm{pF}: 10.80 \mathrm{pF} ; 25.90 \mathrm{pF}$
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| IS4 |

 $\begin{array}{cr}\text { 3x310pF } & \text { 215p* } \\ 003 \times 25 \mathrm{pF} & 435 \mathrm{p} \\ & 43\end{array}$

 
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$3 \mathrm{~A} / 40 \mathrm{~V} \star$
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| :---: | :---: |
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| 20 | ZENERS |
| :---: | :---: |
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## digest.......

## LOST AND FOUND AT SEA DEPARTMENT

An interesting variation on the programmable calculator has sailed into our offices, Texas Instruments have introduced a TI58, complete with brass handled, mahogany case and adaptor/charger. Software in-
cludes a 30 programme navigational package. It will tell you just about everything from where you are, to how fast you'll be going somewhere else. Want to know more, then contaict: Texas Instruments Limited, European Consumer Division, Manton Lane, Bedford MK41 7PA.


## ELECTRONIC SUMIMER SCHOOL

The Department of Electrical Engineering Science at the University of Essex will be holding its annual electronics summer school for teachers during the week 9th-13th July, 1979. This year, as well as courses in linear circuit design and digital circuit design, a third course in electronic systems is also available which is closely related to the A.E.B. electronics systems Alevel. Further information on the Summer School may be obtained from The Department of Elec trical Engineering Science, University of Essex, Wivenhoe Park, Colchester CO4 3SQ.

## ORIENTAL TELETEXT

Sony are to launch a Teletext equipped receiver with infra red remote control. This is the first eastern set in a virtually all-European market. Costing about $£ 800$, it is likely to give the home TV industry some added headaches. The sets are to be built at the Bridgend factory.

## CATALOGUE CORNER

This month's releases include the 1979 Marshalls catalogue, usual comprehensive assortment of components and hardware. Interesting to see they deal in KIM and PET, all in all not bad for value for your 40 p .


## Codespeed electronic mall orider <br> All Full Spec. Devices

TO3 HEAT SINKS!!! Two types of heat sink. Ex equipment, but condition as new Most still contain a power transistor (condition unknown). 'Christmas tree' type. $92 \times 66 \times 35 \mathrm{~mm}$ 20p each. Rectangular type $130 \times 63 \times 32 \mathrm{~mm} 30 \mathrm{peach}$. Please add 25 p per heat sink post and packing.
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PACK 51.25 miniature glass 1 N 3470 germanium diodes ( $600 \mathrm{~mA}, 35 \mathrm{v}$ ). All brand new (at just $2 p$ each how can you go wrong?). 25 diodes for 50 p .
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high-speed switching diodes Same as 1N4148, but ha higher P.IV. 10 for 35p.
PACK 1 . Win to 1 C . we molude the date sheet and PACK E2. Calculator sty high accuracy digital multimeter. Only $£ 3.95$.
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0.33. Win data only $£ 2.95$.

EVER THOUGHT of using 7 segment gas discharge cisplays as an alternative to LED's or LCD's? Gives a nice bright orange display and are comparatively very low in price. Requires 180 v d.c. supply (easily achieved in mains-operated projects). Alf have right-hand decimal points and are supplied with data

PACK E4. 0.3" high $1 \frac{1}{2}$ digit display. Now only 50p.
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## All Untested Packs

PACK M4 CALCULATORS!!! This pack contains a production lime reject calculator Either repair them (not much wrong with some of them) or strip them for spares. Lots of accessible goodies inside, approximately 25 transistors, 2 chips display case and detachable keyboard. Such a bargain, you can't go wreng. Only $\mathbf{E 2} 50$
PACK MU1 (untested - so na quarantees). $2 \times$ Upper half of hand held
case with integral keyboard Ex-equipment but believed to be OK A gift a K Agittat only 50p

PACK DL1 (untested - so no guarantees). A bumper pack of 30 mixed I C 's. You test them and save £ef's. Could include anything linear or digital. A snip at only E3.00.
PACK E1 (80\% guaranteed good). Contains 5 seven segment LED displays. Digit height $0.127^{\prime \prime}$ with right-hand decimal. Common cathode. Still only $£ 1.00$. Your satisfaction is guaranteed or return the complete pack for replacement or a refund.
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## LCD POCKET TV

Matsushita have developed a pocket TV with better resolution than any previous LCD type. 57,600 elements are arranged in a $240 \times 240$ matrix which measures 2.4 inches (presumably diagonally). But even though CMOS circuits are used the TV consumes 1.5 W .

## LANDLUBBER'S CALCULATOR



## BBC TV TRANSMITTER OPENS AT LOCH NESS

We couldn't resist this one, the transmitter is sited at Wester Erchlite, opposite Urquhart Bay. We hear the installation is O.K. so far, but they have had a few teething troubles with programmes like 'All Creatures Great and Small,' something to do with frequency loch.

For those of us who are landbased, and taking $\mathrm{O} / \mathrm{A}$ or degree level studies, Commodore have introduced an updated version of their successful 4190R, designated the SR9190R. It has nine memories with over 100 scientific functions at only $£ 30$. It has all the usual features, $10+2$ LED display, rechargeable batteries and a 1 year guarantee. Your local calc. shop should be able to show you it in action.

## NEW CASES FROM VERO

New Eurocard-sized cardframes have just been announced by Vero. The frames, called the KM6, are available from Vero Electronics Ltd. Industrial Estate, Chandlers End. Eastleigh, Hampshire SO5 3ZR.

# SHEVENON Electronic Components 

## METAL FILM RESISTORS

A range of high precision, very high stability, low noise resistors. Rated at $1 / 4 \mathrm{~W} .1 \%$ tolerance.
Available from 51 ohms to 330 K in E24 series. Any mix

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& & \text { each } & 100+ & 1000+ \\
1 / 4 W & 1 \% & 4 p & 3.5 p & 3.2 p
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Special development pack consisting of 10 of every value from
51 ohms to 330 K (a total of 930 resistors)
£23.75

## BRIDGE RECTIFIERS

| Type | PIV | l |  | Type PIV | I |  |  |
| :--- | ---: | :--- | :--- | :--- | :--- | :--- | :--- |
| W005 | 50 | $1 A$ | $22 p$ | 2 2KBB10 100 | $2 A$ | $39 p$ |  |
| W01 | 100 | $1 A$ | $25 p$ |  | 2KBB20 200 | $2 A$ | $45 p$ |
| W02 | 200 | $1 A$ | $30 p$ | 2KBB40 | 400 | $2 A$ | $50 p$ |
| W04 | 400 | $1 A$ | $35 p$ | BY225 | 200 | $4.2 A$ | $100 p$ |

## REGULATORS

|  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| $78 L 05$ | $30 p$ | $79 L 05$ | $70 p$ | LM309K | 110p |
| $78 L 12$ | $30 p$ | $79 L 12$ | $70 p$ | LM317 | 220p |
| $78 L 15$ | $30 p$ | $79 L 15$ | $70 p$ | LM323K | $530 p$ |
| 7805 | $60 p$ | 7905 | $80 p$ | LM723 | $35 p$ |
| 7812 | $60 p$ | 7912 | $80 p$ |  |  |
| 7815 | $60 p$ | 7915 | $80 p$ |  |  |

## SWITCHES

Subminiature toggle. Rated at 3 A 250 V .
SPDT 65p SPDT centre off 70p DPDT 75p DPDT centre off 90p
Standard toggle.
SPST 34p DPDT, 48p
Wavechange switches.
1 P12W, 2P6W, 3P4W or 4P3W all 37p each.
Miniature switches (non-locking)
Push to make 15p Push to break 20p

## THYRISTORS AND TRIACS

Plastic cased Thyristers. Texas.

|  | 4 A | 8 A | 12 A |
| :--- | :--- | :--- | :--- |
| 100 V | 36 p | 45 p | 62 p |
| 200 V | 42 p | 53 p | 68 p |
| 400 V | 51 p | 66 p | $86 p$ |

Plastic cased Triacs. Texas.
All rated at 400 V

| 4 A | 70 p | 12 A | 90 p | 20 A | 185 p |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 8 A | 80 p | 16 A | 95 p | 25 A | 215 p |

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| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | 2T×300 | $16 p$ |
|  |  |  |  | 2N697 | 12p |
|  |  |  |  | $3 \mathrm{~N}_{1302}$ | 38p |
| AC127 | 17p | BCY71 | 14p | 2N 2905 | 22p |
| AC128 | $16 p$ | BCY72 | 14p | 2N2907 | 22p |
| AC176 | $18 p$ | BD131 | 35p | 2N3053 | 18p |
| AD161 | 38p | BD132 | 35 ${ }^{\text {p }}$ | 2N3055 | 50p |
| AD162 | 38p | BD135 | 38p | 2N3442 | 135p |
| BC107 | 8 p | BD139 | 35p | 2N3702 | 8 p |
| BC108 | 8 p | BD140 | 35p | 2N3704 | 8 p |
| BC109 | 8 p | BF244B | 36p | 2N3705 | 9 p |
| BC147 | 7 p | BFY50 | 15p | 2N3706 | 9 p |
| BC148 | 7 p | BFY51 | 15p | 2N3707 | 9 p |
| BC149 | 8 p | BFY52 | 15p | 2N3708 | 8 p |
| BC158 | 9 p | MJ2955 | 98p | 2N3819 | 22p |
| BC177 | 14p | MPSA06 | 20p | 2N3904 | 8 p |
| BC178 | 14p | MPSA56 | 20p | 2N3905 | 8 p |
| BC179 | 14p | TIP29C | 60p | 2N3906 | 8 p |
| BC182 | 10 p | TIP30C | 70p | 2N405B | 12p |
| BC182L | 10 p | TIP31C | 65p | 2N5457 | 32p |
| BC184 | 10p | TIP32C | 80p | 2N5458 | 30p |
| BC184L | 10 p | 2T×107 | 14p | 2N5459 | 32 p |
| BC212 | 10 p | ZT×108 | 14p | 2N5777 | 50p |
| BC212L | 10p | DIODES |  |  |  |
| BC214 | 10p |  |  |  |  |
| BC214 | 10p | 1 N914 | 4p | 1N4148 | 3 p |
| BC477 | 19p | 1 N4001 | 4 p | 1 N5401 | 13p |
| BC47B | 19p | 1N4002 | 4 p | 1 N5402 | 15p |
| BC479 | 19p | 1 N4004 | $5 p$ | 1 N5404 | 16p |
| BC548 | 10 p | 1 N4006 | 6 p | 1N5406 | 18p |
| BCY70 | 14p | BZY88 series 2 V 7 to 33 V 8 p each. |  |  |  |
| LNEAR A SELECTION ONLY! DETAILS IN CATALOGUE. |  |  |  |  |  |
| 709 | 25p | LM324 | 50p | NE556 | 60p |
| 741 | 22p | LM339 | 50p | NE565 | 120p |
| 747 | 50 p | LM380 | $75 p$ | NE567 | 170p |
| 748 | 300 | LM382 | 120p | SN76003 | 200p |
| CA3046 | 55p | LM1830 | 150p | SN76013 | 140p |
| CA3080 | 70p | LM3900 | 50p | SN76023 | 140p |
| CA3130 | 90p | LM3909 | 60p | SN76033 | 200p |
| CA3140 | 70p | MC1496 | 60p | TBA800 | 70p |
| LM301AN | 2Bp | MC1458 | 35p | TDA1022 | 650p |
| LM318N | 125p | NE555 | 25p | ZN414 | 75 p |

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## LEDs $0.125 \mathrm{in} . \quad 0.2 \mathrm{in}$.

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13 p
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0.068, 0.1
$3 p$
4 p
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| C04002 | 0.17 | ここ－029 | 1.03 | C04053 | 0.82 | CO4093 | 0.80 | CD40193 | 1.40 |
| C04006 | 1.04 | ここ．0こ0 | 0.50 | CD4054 | 1.04 | CD4094 | 1.69 | CD40194 | 1.19 |
| C04007 | 0.18 | ここと込 | 2.00 | CD4055 | 1.18 | CO4095 | 0.94 | CD40257 | 1.48 |
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| c－：\％ | 0.43 | CD4037 | 0.85 | CD4067 | 3.35 | CO40101 | 1.61 | CO4516 | 1.01 |
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## FIBRE OPTIC LIGHT

## PEN

Light pens have never figured very greatly in the amateur mar－ ket，but this device from Optronic Fort Ltd hopes to change that， particularly with the tremendous interest in mini－computers．It is TTL compatible and uses a pin－ photodiode，weighs only 35 grammes and can be yours for only $£ 175$ ．Call them at Cam－ bridge Science Park，Milton Road，Cambridge CB4 5BH．

## VIDEO DISC 2

Further to the Philips video disc launch，news has just arrived about the RCA＇Selecta Vision＇ video disc system，pioneered in the USA．The RCA system uses a grooved disc，with diamond stylus．The disc rotates at 450 rpm，and has one hour＇s playing time per side．Again only draw－ back is you can＇t record your own material，but just think，you will be able to buy your own copy of Star Wars or even Emmanuel， if your that way inclined，and you can see the good bits over and over again，Cor．

## BREADBOARD＇79

What！already．Well this year there are two dates to put into your Letts Electronic diaries．The Midlands show will be at Bingley Hall，Birmingham，on May 23rd－ 26 th，and the London show is at the Royal Horticultural Hall， Westminster，December 4th－8th． Figures show over 10000 people attended the first show，and that＇s a fantastic response for the first ever home electronics exhibition，indeed the Birmingham show is due entirely to response from contributors and visitors，we＇ll see you there．

## SINCLAIR AGAIN

Two new laboratory quality mul－ timeters are promisedior 1979 As usual you can expect the Sinclair innovations in cost and features． These instruments which rejoice
under the titles of DM450（ $41 / 2$ digit 5 function）and DM350（ $31 / 2$ digit， 34 ！ranges）．They both have good technical specifications and accessories．See the Microvision article for Sinclair＇s address．

## MEMORIAL FUND

Following the tragic death of John Miller Kirkpatrik a memorial fund has been esta－ blished for the benefit of his family．Those wishing to make a donation should send this to： John Miller Kirkpatrik Memorial，Lloyds Bank Ltd， 39 Threadneedle Street，London EC2．


## FAST AND EASY FASTANDEASY FASTAND EASY FAST ANDEASY



If you continually solder and de-solder when building, testing and trying out modifications. on circuits, you're wasting money. Heat damage and solder build-up waste boards and components faster than anything.

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Breadboards are fully re-usable and avoid these losses - you only need to hard wire your circuit to keep the final design.

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Your components - IC's or discretes - simply plug in. Your interconnections are merely a matter of pushing in solid wire - up to 0.032 inches.

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# DIGITAL TO ANALOGUE TECHNIQUES 

## Digital to Analogue conversion (DAC) is a fast growing section of electronics. Tim Orr explains some of the more practical applications.

ELECTRONICS HAS CHANGED enormously in the past ten years, having swung away from valves, germanium transistors, even from descrete devices themselves. The trend is towards more and more complex integrated circuits, complete systems in a chip, large scale integration (LSI). Also the trend has swung heavily towards digitally based systems rather than analogue ones, partly because the IC manufacturers can get a greater success rate from making digital devices and partly because there are very many applications which can only be contemplated with a digital device. Such examples as pocket calculators and microprocessors spring immediately to mind. However there are several areas where analogue techniques present the only realistic solution (at this moment in time), such as tone controls in an audio amplifier. In fact, good cases can be made out for both analogue and digital systems and there are many examples where both are needed. In these it will be necessary to change from the analogue to the digital world or vice versa and to do this, some sort of conversion process has to be practised.

## Digital to Analogue Conversion.

The job of a digital to analogue converter (DAC) is to convert a binary code (a digital data word) into an analogue voltage. The data word is a digital representation of that analogue voltage. Thus if we presented the DAC with a digital word that was linearly increasing in magnitude, the output would be a linearly increasing analogue voltage. This digital word would be the output of a binary counter driven by a constant clock frequency. The analogue output is a linear ramp, or rather a linear staircase where the step size is controlled by the "size" of the DAC. If the DAC is an 8 bit device, ie it can accept data words 8 bits wide, then it can generate a possible $2^{8}$
descrete output level. Now $2^{8}$ is 256 , so therefore an 8 bit DAC could generate a staircase with 256 steps in it. The resolution of the DAC is thus 1 part in 256 , or rather a change of one LSB (least significant bit) in the data word will make the output voltage change by $1 / 256$ th of the full scale output.

To get really fine resolution then a high performance DAC is needed. DAC prices seem to be almost linearly proportional to their resolution. I have got several DAC's amongst my collection of bits. There is an 8 bit DAC costing about $£ 4$, a 12 bit DAC costing about $£ 35$ and a 16 bit DAC costing just over $£ 200$. It is now possible to buy a monolithic (a single I C') DAC with a bit size of 6 , 8,10 and 12 , but above this the devices are usually modular.

## Size And Resolve

Fig 2 shows the relationship between DAC size and resolution. Notice that a 16 bit DAC with a 10 V full scale output is made up of a staggering 65,536 descrete



Fig 1. Converting binary code to analogue voltage.

| WORDLENGTH n | RESOLUTION 1 PART IN 2n | MAXIMUM THEORETICAL DYNAMIC RANGE | BIT SIZE <br> ASSUMING <br> FULL SCALE $=10 \mathrm{~V}$ |
| :---: | :---: | :---: | :---: |
| 1 | 2 | 6dB | 5.0 V |
| 2 | 4 | 12 dB | 2.5 V |
| 3 | 8 | 18dB | 1.25 V |
| 4 | 16 | 24dB | 0.625 V |
| 5 | 32 | 30dB | 0.312 V |
| 6 | 64 | 36dB | 0.156 V |
| 7 | 128 | 42dB | 78.1 mV |
| 8 | 256 | 48dB | 39.1 mV |
| 9 | 512 | 54 dB | 19.5 mV |
| 10 | 1024 | 60 dB | 9.7 mV |
| 11 | 2048 | 66 dB | 4.8 mV |
| 12 | 4096 | 72dB | 2.4 mV |
| 13 | 8192 | 78dB | 1.2 mV |
| 14 | 16384 | 84dB | 610 uV |
| 15 | 32768 | 90 dB | 305 uV |
| 16 | 65536 | 96 dB | 152uV |

Fig 2. Relationship between size and resolution.
levels each $152 \mu \mathrm{~V}$ in size. (There is also available an 18 bit device, costing a small fortune. The larger the bit size of the DAC, the larger is the dynamic range (best signal to noise ratio) of its output. This increases by 6 dB per bit. Thus a 10 bit DAC can give a best range of 60 dB .

The human anatomy has developed over the last few million years to respond to its environment. This has resulted in the following performance figures. The sensitivity of the eye to colour is not that good. Colour television transmission doesn't give much of its bandwidth to the colour part of the signal. Have a look at a TV and see how well defined the colour is; it is usually just "sort of smeared around' the subject. Thus it is possible to get quite good digital video using only 4 bits for the colour. The eye sensitivity to resolution is somewhat better, but even so an 8 bit oscilloscope memory will look fairly continuous, giving little indication that it is made up of descrete steps.

## Ear Lead

However the ear can still outperform present day technology. Using a 16 bit high quality audio system a trained ear can still detect the difference between the digitally processed sound and the original. Thus, when using DAC's in professional audio equipment great care has to be taken to eliminate all types of abberations in the system. These digital abberations don't just worsen
the signal to noise ratio as an analogue system might), but they produce discordant harmonic distortion, sidebands like those obtained from ring modulation and other little funnies.

Figure 3 shows a DAC system in operation. The output of the DAC is meant to produce nice clean square wave steps, but the leading edges of these steps always have small spikes (glitches), caused by the switching times associated with the DAC's internal workings. These glitches are not regular in nature and so filtering cannot eliminate them. The glitches give the sound a "dirty" quality, or, if the system is an oscilloscope display it produces fuzzy pictures.

The glitches can be removed with a little module called a DEGLITCHER, fig 4. This is a logic controlled sample and hold which holds during the glitch period, but otherwise tracks the signal from the DAC. Thus the glitches are ignored. The output from the deglitcher then passes through a low pass filter and this removes the "stepped" quality of the signal and produces a smooth analogue output. The cut off frequency of this filter is very important and is related to the data rate of the DAC, The rule of thumb is that the filter cut off frequency should always be less than half of the data rate frequency

## Buying And Building

DAC's can be bought fairly cheaply as complete IC's or they can be constructed out of generally available parts, fig 5. This circuit uses precision buffers (a CD4041 will do), E24 resistors and a FET op amp. The buffers are run from $a+10 \mathrm{~V}$ supply and their purpose is to provide high $(+10 \mathrm{~V})$ and low ( 0 V ) output with low source resistance. They are driven by a 6 bit data word, the MSB (most significant bit) thus drives the 7 k 5 resistor, the LSB (least significant bit) the 240 k resistor. So, when the MSB changes, the output of the op amp will move by a large amount 5 V ), but when the LSB changes the output will only change a little (OV156). Going from the MSB down to the LSB, each bit has only half the effect of its predecessor. This is obtained by doubling the resistor values ( $7 \mathrm{k} 5,15 \mathrm{k}, 30 \mathrm{k}, 60 \mathrm{k}, 120 \mathrm{k}, 240 \mathrm{k}$ ).

A 6 bit DAC can produce $2^{6}$ descrete output levels. Now $2^{6}$ is 64 and so the overall resistor tolerance should be +1 part in $2 \times 64$, which comes out at $\pm 0.8 \%$. This type of DAC is known as a resistance ladder DAC, but in its presented form it is rather limited. For instance, a 10 bit device would require a resistor range of 1024 to 1 and a tolerance of $0.05 \%$.



Fig 6. An R-2R ladder.
still applies. Also the ratio between the resistor value and the buffer ON/OFF resistance is important. The $2 R$ resistors connected to the buffers should ideally be $2 R-$ (the buffer output resistance).

## Counting On This

A "counting" type ADC is composed of a fast comparator, a gate, a counter and a DAC. This is why ADC's always cost more then DAC's, the ADC uses a DAC to do the conversion. Assuming that the analog input is positive, and the DAC produces a positive output, the conversion operation is as follows:

1) The signal "stan conversion" is generated. This resets the counter to all zero's, the DAC output goes to zero, the comparator output goes high and so the clock is allowed to enter the counter. Thus the count proceeds and the DAC generates e positive going staircase.
2) When the DAC output exceeds the level of the anaiog input the comparator output goes low, the counter stops. This is the end of the conversion, and the date that is held on the counters output is the data output. It would then be transferred to some latches, and held there until the next conversion is finished.

This data word describes as precisely as is possible the magnitude of the analogue input. Although simple to operate, this method has a major disadvantage, it is slow. Imagine that the ADC is a 10 bit device and the clock frequency is 500 KHz ,

then the longest conversion time will be 1024 counts at $2 \mu$ Sec per count which is 2.048 mSec , this means that the conversion rate will be less than $\mathbf{5 0 0}$ per second.

## Memory Planning



The data that drives DAC's can come from several sources. It could be generated by computation or read from a programmed memory as shown. In this example a ROM (read only memory), hes been programmed with the data necessary to produce a
sinewave. An updown counter provides the address for the ROM and the data is converted into an analog output by the DAC. The clock frequency divided by the size of the counter determines the sinewave frequency.


## Data Lining

Another method of generating data is to convert analogue information into digital words. The signal must first be passed through a low pass filter, the cut off frequency of which must be less than half of the conversion frequency. The signal is then "held" in a sample and hold unit so that the ADC can do its conversion on a static signal. Control logic sends commands to the ADC giving it various instructions. The sequence of events is:

1) Tell sample and hold to HOLD.
2) Tell ADC to start conversion (SC).
3) Conversion finished, generate end of conversion signal (EOC).
4) Tell sample and hoid to SAMPLE.

The process then repeate itself. The sample and hold mechanism is shown below. Generaliy, in one period and the input signal several ADC conversions will be done. The data generated is then stored, processed or transmitted

## Mark Time

FILTERED OUTPUT

MARK SPACE SIGNAL
FROM SWITCH

Yet another type of DAC, a mark space modulation DAC is shown above. The data word is presented to one side of a magnitude comparator, the output from a fast running counter to the other. When the counter is greater than the data word the $A>B$ output goes low. The output is a mark space waveform the ratio of which is linearly proportional to the magnitude of the
data word. The mark space signal operates a precision switch, the output of which is lowpass filtered, providing a smoothed DC output. This type of DAC requires a fast running counter, but gives a relatively low bandwidth output signal. It is a good solution for a system where lots of slow moving outputs are required, because the counter can be common to all the DAC's.

Digital Echo Chamber


There are several professional echo chambers that are all digital.' The audio input is converted into a digital word and then put into a parallel set of shift registers. A 10 bit system would use 10 sets of registers. The clock that starts the ADC conversion also shifts the data along the shift registers. The data coming out of the shift registers is then converted back into an analogue voltage by the DAC. It is then filtered and mixed with the original signal.

The echo can be made to repeat indefinitely by using the digital recirculate path around the shift registers. The amount of
digital storage required is rather large. Let us assume that we want a good quality echo. This would be a 10 kHz bandwidth, 60 dB dynamic range which implies a clockrate of about 25 kHz and a 10 bit system. Thus to store 1 second of sound (to give one second delay), we would need $10 \times 25,000$ bits of memory, 0.25 Mbits!

The usual solution to this dilemma is to get longer delays at the expense of bandwidth. Thus a 1 second delay would be 1 kHz bandwidth, a 0.1 second delay would be 10 kHz bandwidth. Th is would only require 25 K of memory.

## Multiplexed Sound System

Next time you are on an aircraft with a multichannel music system, it is quite possible that the sound you are hearing via your stethoscope is digitally generated. The sounds are usually stored on a multichannel tape player and each channel is connected to a multiplexer. This is a digitally controlled rotary switch and it is continually scanning all the audio channels. The output of the multiplexer is then fed to the ADC. Thus each channel is converted to a digital code. This digital code is then transmitted in serial mode and mixed with a sync pulse. The transmitted information is a series of serial data words, each representing a small piece of the eight music channels, plus some synchronisation data which passes down a two wire system to each receiving unit. This saves wire weight, there is less crosstalk and low pickup due to the high noise immunity of digital systems.


## Laser Light Show.

One of the recent laser light shows in London used a digital tape recorder to store the data for the show. Two outputs were produced which were converted into control voltages by DAC's. These voltages were then used to manipulate the $X$ and $Y$ Co-ordinates of the laser. Thus it was possible to draw pictures and cartoon characters with a moving laser beam.


## Digital Memory for an Oscilloscope.

There are several products on the market that enable an ordinary oscilloscope to store waveform information. This is particularly useful if you are trying to capture non-repeating events. The system is very similar to the digital echo unit, there is an ADC, a memory and a DAC. Also there is a trigger circuit so that one shot events can be captured and a ramp generator to produce the Xsweep. The output of the DAC is rather interesting, because it is not low pass filtered, but it uses a linear point interpolation device. Basically, what this does is to join up the dots, so that a waveform that is represented by only a few points, can be made to look like the original signal. The visual results of interpolation are very good indeed.


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A light chaser is a mechanical or electronic gadget which controls three or more lights arranged in a chain; these are flashed on, one at a time, in sequence to create an illusion of movement. They are used at fairgrounds, in advertising, in shop windows and in discos. Our project to build one is both simple and easy to build.

Decibels


Not surprisingly those who are new to electronics are confused by the apparently crazy use of decibels to describe gain or attenuation. Why not use easily understood numbers? We tell you and hope to convert you.

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A project for those of you who do more than click the shutter. Our unit is in the mains lead to your enlarger (although battery operated) and allows you to set exposure times between 0.9 and 100 seconds in two infinitely variable ranges.

Communications Satellites


Speak to someone on the phone outside Europe and the chances are that your voice will spring out into space for thousands of miles on the way. The commercial ends of the space programme are described.

Telephones


Do you know how the phone, one of the most widespread pieces of electronics, works? Lots of exciting things are happening on this front; we pull back the curtain and take a peep next month.

Crossing your Bridges


The Wheatstone Bridge is one of the commonest circuit configurations in electronics. Next month K. T. Wilson examines the theory of this and describes the variations that we now use.

Experimenters Power Supply
Second in our series of test gear projects is a $0-20 \mathrm{~V}, 1 \mathrm{~A}$ bench power supply, stabilised of course as well as short circuit protected.

Workshop Test Gear
The HE project team have prepared a feature giving their views about what you need in the way of test gear in your workshop. It's a thoroughly practical approach and continually bears in mind the limitations of finance.

How TV Signals are Propogated


Put up an aerial in most areas of Britain and you'll have no trouble in getting a good signal but that's only because the broadcast engineers have taken into account a multitude of factors. We take a look at this subject in the March issue.

The March issue will be on sale February 9th
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# HEADLIGHT DELAY 

## Use your car headlights to give post-parking illumination with this simple unit.

THIS SIMPLE LITTLE UNIT lets you use your car head or spot lights to illuminate your pathway for a pre-set period of about 50 seconds after you have parked the vehicle. At the end of this period the unit turns the lights off automatically

The unit thus enables you to avoid walking into dustbins or tripping over junk that may be obstructing your private driveway, and helps you avoid stepping into various nasty bits that may be laying on the public sidewalk. The unit is easy to install in the vehicle.

## Construction and Use

Construction of the unit should present no problems at all. The relay can be any 12 V type with a coil resistance of 120 ohms or greater, and with two or more sets of N.O. contacts that are rated at 3 amps or greater.

When it comes to installing the unit, note that two methods of correction to the vehicle are possible. On some vehicles the headlight switch is connected directly to he battery so that the headlights operate even when the ignition is turned off (see Fig 2a). In this case take connection 4 of the 5 -day terminal block directly to the live side of headlamp switch SW 1, and connection 5 to the headlamp side of SW1

The alternative connecion is shown in Fig 2b. Here, the headlight switch is wired in series with the vehicles ignition switch, so that the headlights only operate when the ignition is turned on. If your vehicle uses this type of connection, take connection 4 of the 5 -way terminal block to the live side of the ignition switch, and take connection 5 to the headlamp side of SW1.

## BUYLINES

With the small number of components involved, it would be surprising if there were any problems in obtaining them.


Fig 1. (Below) Circuit diagram of the unit
HEADLAMP SWITCH


HOW IT WORKS

The unit is designed around a type- 555 timer i.e., with a relay output. The relay has two sets of normally-open contacts. Normally, START switch PB1 and the relay contacts are open, so zero power is fed to the timer circuit and (assuming that HEADLIGHT switch SW1 is open) the headlights are off. Circuit Action is initiated by briefly closing push-button switch PB1. When PB1 is momentarily closed power is fed directly to the relay coil, and the relay turns on. As the relay turns on contacts RLA/ 2 close and apply power to the headlights and contacts RLA/1 close and apply power to the timer circuit, but pin 2 of the IC is briefly tied to ground via C1 and R3 at this moment, so a negative trigger
pulse is immediately fed to pin 2 of the IC and a timing cycle is initiated. Consequently, pin 3 of the IC switches high at the moment that the relay contacts close, and tbus locks the relay on irrespective of the subsequent state of switch PB1.

The 555 is wired as a one-shot timer or monostable with a timing period of about 50 seconds (determined by R2 and C2). Thus, the relay and headlights are held on for the duration of this 50 second timing period. At the end of the timing period pin 3 of the IC switches to the low state, so the relay turns off and contacts RLA/l and RLA/ $/ 2$ open, removing power from the timing circuit and the headlights. The operating sequence is then complete.


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## Cellular LoGic Mage mocessing


#### Abstract

At University College, London, there is a research group working on a method of image processing which could prove to be the link between the human eye and the TV camera. Computing Today's Phil Cohen talked to Dr. Michael Duff about Cellular Logic Image Processing.


CELLULAR LOGIC IMAGE PROCESSING was first proposed in 1958 by $S$. Ungar in the States. It was suggested that the cells of the human eye do a lot of the processing before what we see is fed up the optic nerve to the brain.

What exactly do we mean by image processing?
Generally, it means processes like perimeter-finding - producing the outline of an object, or skeletonising finding a set of lines which are unbroken and follow the object's shape.

This sort of process can be used in such diverse applications as fingerprinting, character recognition (OCR) or even intruder detection (spotting movement on a TV picture) but perhaps the two most useful areas will be biomedical scanning - chromosome counting or looking for abnormalities on X-ray plates - and production line quality control.

## Parallel Processing

The model of the human eye previous to 1958 was of a simple camera - the point-by-point information was fed to the brain, which did all the clever processing.

However, it was pointed out that for processes such as edge-finding it was much more efficient to use a
parallel processing system
The essential difference between serial and parallel: processing schemes is that in a serial scheme the data is processed bit by bit in a central unit (CPU) and the intermediate results are stored in memory. In parallel processing the data is fed in as an array and the processing takes place all at the same time - there is one processor for each data element. The intermediate results are passed from processor to processor as the calculations continue.

In the human eye, then, the question is: could a number of cells just behind the light-detecting ones be the parallel processors, responding to commands from the brain to find the edges of objects, or detect movement? Certainly it is known that the edges of the field of vision are extraordinarily good at spotting movement could this be because the structure of the eye is different there?

## The Processors

Going back to the CLIP machine, in this sort of application the type of processor we are talking about is in no way as complex as a modern MPU. The sort of data it receives are single-bit inputs from the image sensor

8 INPUTS FROM NEIGHBOURING CELLS

PROCESSED
PATTERN
DISPLAYS

COMMON OUTPUT
TO \& NEIGHBOURING CELLS

Fig. 1: The block diagram of one of the processors.


Fig. 2: Stages in the processing of a microscope picture of part of a rat's anatomy.
associated with it and one bit each from the eight nearest processors' outputs.

Why eight? Well, this provides an optimum "connectivity" - too few and the processing becomes slow, too many and the cost of connecting the processors together becomes enormous.

The sort of operation the processor would have to perform would be to give an output if any of its neighbours gave an output and the image bit fed to it was a " 0 ".

The program example (in PET BASIC) given shows the usefulness of this sort of process. Of course, we cannot perform parallel processing directly in BASIC the program has to scan the image bit by bit, simulating the action of each processor in turn.

The important thing about using this sort of scheme for image processing is that the outputs of the units will change in "waves", travelling at speeds dependant on the propogation delay of the devices involved. This means that, by having four "edge registers" which are not connected to the image input, we can do things like finding the outer edge of an object by starting a signal from the edge registers and programming the processors to stop propogating it at the edge of the object. The program example carries out this sort of process.

## Structure

In the CLIP machine, the processors each have the structure shown in Fig 1. Each is connected to its eight neighbours and its output fans out to the same neighbours. There is also a "pattern input" for connection to the picture signal (which is derived from a TV camera and multiplexed to provide each processor with a 1-bit signal from one point of the camera's image).

The gate enabling threshold selection and function control inputs are from a programming bus common to all processors.

The gate enabling inputs allow instructions like "If the output from the processor to the left is ' 1 ' . . . "The threshold selection inputs allow "If more than three inputs are ' 1 '. . .'

Combining the two allows very comprehensive processing of the inputs - "If any two of the processors to the left give and output . . ." for example.

There are also various buffers for more complex instruction types.

The boolean processor itself can be programmed via the function controls to "look" like any combination of memory-less logic gates.

## Implementation

The processors come in custom-built ICs, each chip containing eight units. The CLIP 4 machine contains an array of $96 \times 96$ processors.

CLIP 4 is the product of ten years of research at University College. It's a commercially viable product it fits into one 7 -foot instrumentation rack, including power supplies and controller. The cost? In the region of £ $30-40$ thousand.

The processors themselves are based on NMOS technology and the control circuitry (the part that acts as a "conductor" - in the musical sense - directing all of the processors) is implemented in hardware - an MPU would be too slow!

The input signal is from a TV camera (only part of the picture is used - 96 lines $\times 96$ points). This is encoded either as a black and white picture with no grey or as a grey-scale image. CLIP can handle grey-scale pictures, performing processes such as smoothing.

The output from the system would be to a video monitor or, in some applications, just a few bits of data to another peripheral, such as a warning indicator in the case of intruder detection.

## Software

The software for the system consists of a series of inputs for the function-definition bus of the processors and a loop structure which is linked to the processor outputs.

Looking at Fig. 2, the machine is trying to find the outline of the largest isolated mass of black in the input pattern.


One of the chips in the CLIP - each contains eight processors.
The original input is shown in the top left. The first instruction propagates white from the edge registers through all connected black. This leaves the pattern second from the left on the top line. The program then
'erodes' the image by removing all black dots not surrounded by black and then removing their neighbour black dots as well.

It repeats this erosion until one more step would cause all black to vanish completely. This leaves the image as it is at the end of the top line.

The program then surrounds each black by eight blacks. It does this twice. It then recalls from the original input pattern the part which is "connected" to the current pattern. The last step finds its outline.
.Naturally, this sort of software cannot be written in a conventional language - the group have developed what is effectively an assembler for the system and all the groups working on image processing worldwide are due to meet this spring to discuss a suitable high-level language.

## Applications

One very interesting application mentioned earlier is production line control. CLIP can tell the difference between an object which has been correctly punched out of metal and one with the wrong surface area or the wrong number of holes, etc.

The amazing thing is that it can do this fifty times a second! In fact, the machine can perform 1500 parallel processes per TV frame period.

The machine could be fitted to the "reject" solenoid on a production line so that badly produced pieces could be pushed off the line.

Another area in which the machine could be useful is in microscopic counting. There are systems available already which will count the number of items in a picture, or even the number between certain size limits, but the inherent flexibility of CLIP make it invaluable for complex tasks such as red blood cell deformity checking and other applications where previously a human operator was the only alternative.

One slightly more frightening possibility is the use of such a system in facial recognition - enabling authorities to keep track of every individual automatically.

When the system was first proposed about ten years ago, the device which was envisaged was a pair of super-binoculars, with photo-diodes at one end and LEDs at the other, modifying images so that only moving objects, or even more selectively, only enemy tanks would be seen! This is some way from the present state of the art but in a few years . . . who knows?

We would like to thank Dr Michael Duff and University College in general for their help. ETI

## CLIP SIMULATION PROGRAM

The following program simulates the action of the CLIP machine by pretending to be each processor in turn in a $10 \times 10$ array. It's very slow to run (several minutes) and this shows the advantage which a parallel processing system has over a serial one.
$10 \mathrm{~S}=10$
$S$ is the dimension of the 2-dimensional square processor array 20 DIM A(S, S), B(S, S)
$A$ is the image input to the system. $B$ represents the processor outputs. Load the image into the system:
30 FOR I $=2$ TO S -1
$40 \mathrm{FOR} \mathrm{J}=2 \mathrm{TOS}-1$
50 READ A(S, S)
60 NEXT J
70 NEXT I
The outer layer of processors represent the edge register, in which we can initialise processing 'ripples' (see text).

80 DATA $0,0,0,0,0,0,0,0$
90 DATA $0,1,1,1,1,1,1,0$
100 DATA $0,1,0,0,1,0,0,0$
110 DATA $0,1,1,0,1,1,0,0$
120 DATA 0,1,1,0,1,1,1,0
130 DATA $0,1,1,1,1,1,0,0$
140 DATA $0,0,0,0,0,0,0,0$
150 DATA $0,0,0,0,0,0,0,0$
Now for the 'seed' which will propogate during processing. Note that it's in the edge register:

$$
1010 \mathrm{~B}(\mathrm{~S}, \mathrm{~S})=1
$$

Now print the results so far:
1014 GOSUB 2000
$1015 \mathrm{~F}=0$
$F$ is set to 1 if any changes are made.
1020 FOR I $=2$ TO S -1
1030 FOR $J=2$ TO $S-1$
. . For each processor
1040 FOR K $=-1$ TO 1
1050 FOR L $=-1$ TO l
... For each of the eight 'connected' processors
1055 IF L $=0$ AND K $=0$ THEN 1090
. . Except the one we're simulating
1060 IF $\mathrm{B}(\mathrm{I}+\mathrm{K}, \mathrm{J}+\mathrm{L})<>1$ OR A(I, J) $<>0$ THEN 1090
skips the next bit unless the image is zero at this point and one of the neighbours outputs is one.

1070 IF $B(1, J)=0$ THEN $F=1$
$B(I, J)$ is going to be set to $1 . F$ is set to 1 if this represents a change.

## $1080 \mathrm{~B}(\mathrm{I}, \mathrm{J})=1$

1090 NEXT L:NEXT K:NEXT J:NEXT I
1130 IF $F=1$ THEN 1014
1140 STOP
repeats the process until the output is stable (ie there were no changes during this pass).

The following subroutine prints the results:
2000 REM PRINT
2010 PRINT " ": REM CLEAR SCREEN CHARACTER
2020 FOR I $=1$ TO S
2030 FOR J = 1 TOS
2040 IF A(I,J) $=1$ THEN PRINT "A";:GOTO 2060
2050 PRINT" ":
2060 NEXT J
2070 PRINT " ";
2080 FOR J = 1 TO S
2090 IF B(I,J) = 1 THEN PRINT "B";;GOTO 2110
2100 PRINT " ";
2110 NEXT J
2120 PRINT
2120 NEXT I
2140 RETURN


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|  |  | Jan 77 |
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# LOGIC TRIGGER 

## No need to be in the dark any more, see your pulses, doing their stuff before your very eyes.

WHEN USING AN oscilloscope to examine or fault find digital circuitry, it is often desirable to see what happens just before a pulse or edge occurs. An example of this is when measuring the propagation delay in a ripple counter. Here it is easy to trigger on the last output but the edge of the counter input which initiated the change in the output may have occurred over 100 ns earlier. Even with the delay line built into modern oscilloscopes the edge is too early to see.

Triggering on the input waveform allows this edge to be seen but if the output pulse occurs only once every thousand or so pulses it will not be seen. With this unit, the output of all the stages in the divider can be examined and a pulse can be generated anywhere in the cycle. By selecting a pulse very close to, but before, the edge in question and using it to trigger the oscilloscope (use ext trigger) both the clock waveform and output
 waveform can be see $n$.

## SPECIFICATION

| Modes | Asynchronous or synchronous |
| :---: | :---: |
| No. of inputs | 12 address, 1 clock |
| Loading address clock | $\begin{aligned} & 0.4 \text { UL (TTL) } \\ & 0.4 \text { UL (TTL) } \end{aligned}$ |
| Pulse extension mono | 10 ms |
| Pulse indication | LED |
| Minimum puise detectable | $<40 \mathrm{~ns}$ |
| Propagation delay | $<45 \mathrm{~ns}$ |
| Trigger (synchronous) | positive or negative edge of clock input |
| Set up time (synchronous) address to clock | $<40 \mathrm{~ns}$ |
| Output | logical " 1 " when input agrees with switch setting and/or clock (synchronous only') |
| Power requriement | +5V@ 50 mA |

With the advent of microprocessors it has become increasingly difficult to fault find as things happen (e.g. the CE input to a memory may go low) only when a particular address is given. As the address bus is always in motion it is almost impossible to trigger the scope on any one address. Again with this unit the address bus is interrogated along with the necessary write or read lines, and its output can be used to trigger the oscilloscope only when the correct sequencer is received.

## Construction

We mounted all the components on the board including the switches. The only difficult (fiddly) bit is the writing of the three position slide switches which have to be preassembled before fitting to the pcb. The wiring is shown in fig. 3 .

To aid this we have provided 12 holes in the pcb the size of the toggle of the switches; if the


Fig 1. Full circuit diagram of the logic unit.

## PARTS LIST

| RESISTORS | all $1 / 2 \mathrm{~W}, 5 \%$ |
| :---: | :---: |
| R1-R12 | 47k |
| R13 | 39k |
| R. 4 | 180R |
| R15 | 1k |
| R: 6 | 27k |
| CAPACITGRS |  |
| Ei. 2 | $4 \mu 725$ Velectro |
| C3-C5 | $33 \mu 16 \mathrm{~V}$ tantalum |
| SEMICONDUCTORS |  |
| IC1-IC3 | 74LS86 |
| IC4-IC6 | 74LS02 |
| IC7 | $74 \mathrm{LS30}$ |
| 1 C 8 | 74LS74 |
| LED1 | Red LED |

## MISCELLANEOUS

PC board ETI 141
Twelve 3 position slide switches
Two 2 position slide switches
Front panel
Box to suit

Fig 2 (right): Foil pattern shown full size.

HOW IT WORKS

The twelve inputs are compared to the levels set on the slide switches SW1-SW12 by the exclusive OR gates IC1-IC3. These ICs have a high output only if the two inputs differ. If they are the same, either both low or both high, the output will be low. If the two inputs are joined together, as when the switches are in the don't care position, the output will always be low.

The outputs from the exclusive OR gates are combined in pairs by the NOR gates IC4-IC6. If the 12 input signals match the preset selection, the output of all 6 NOR gates will be high. If any one is not in agreement with the selection one or more of the NOR gates will have a low output.

These NOR gate outputs are combined by IC7 which is an eight input NAND gate. The output of this gate will low only if all 12 inputs match. The output of this IC is inverted by IC4/d to provide the asynchronous output.

This output also triggers the monostable formed by IC6/c and IC6/d. This gives a 10 ms long pulse of light the LED indicationg a pulse was received. If it is a steady state signal the LED will stay on.

The output of the NAND gate, IC7, also joins the data input of IC8 (D type flip flop). This IC is toggled on the positive edge of the clock waveform transferring the data to the output. This is the synchronous output. To allow for either positive or negative synchronization an inverter is used on the clock input and either polarity can be selected by SW13.


Fig. 3 (Below) Component overlay of the logic trigger

switches are initially placed upside down in these holes the board will act as a template to provide the correct spacing. We have also used two wires of the switch to provide mechanical support. While only a single pole switch is needed the
only ones readily available are two pole.

The switches can now be mated to the PC board with the two longitudinal wires being terminated in the holes provided at the end of the switch bank.


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[^1]| Commands |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| CONT | LIST | NEW | NULL | RUN |  |
| Statements |  |  |  |  |  |
| CLEAR | DATA | DEF | DIM | END | FOR |
| GOTO | GOSUB | IF...GOTO | IF...THEN | INPUT | LET |
| NEXT | ON...GOTO | ON...GOSUB | POKE | PRINT | READ |
| REM | RESTORE | RETURN | STOP |  |  |
| Expressions |  |  |  |  |  |
|  |  |  |  |  |  |
| $-,+, *, /, \uparrow$, NOT, AND, OR, $>,<,<>,>=,<==,=10^{-32}$ to $10^{+32}$ |  |  |  |  |  |
| Functions |  |  |  |  |  |
| ABS ( ) $^{\text {( }}$ | ATN(X) | $\cos (X)$ | EXP( X$)$ | FRE( $X$ ) | INT(X) |
| LOG(X) | PEEK(I) | POS(I) | RND( $X$ ) | SGN(X) | $\operatorname{SIN}(X)$ |
| SPC(1) | SQR(X) | TAB( 1 ) | TAN(X) | USR(I) |  |
| String Functions |  |  |  |  |  |
| ASC(X\$) | CHRS(l) | FRE(X\$) | LEFT\$(X\$, 1 ) | LEN(X\$) | MIDS |
| RIGHT\$(X | ( ${ }^{\text {, }}$ ) |  | STR\$(X) |  | (X\$, I, J) VAL(X\$) |

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# WHO NEEDS ELECTRONICS? 

## K. T. Wilson explores the all too frequently ignored and misunderstood field of Magnetic Amplifiers.

THINK OF AMPLIFICATION, and you automatically think of transistors. Perhaps if you're a bit longer in the tooth you remember valves. Have you ever thought of large amounts of power gain being obtained without using either transistors or valves? It's power gain we're talking about, too, not just voltage gain. A transformer will give voltage gain, up to 100 times, but at the expense of current, so that the power out is never quite as much as the power in. There's no power gain there, but a device called the magnetic amplifier, which looks very like a transformer, can give very large values of power gain, can control AC power into a load very smoothly, and is used in the sort of applications where thyristors would be a-natural choice for many.

The magnetic amplifier has been used in industrial control for decades, yet has never really caused any stir of interest anywhere else. Perhaps it's because it's always a ready-made item, but then so is an IC amplifier, and everyone seems to make use of those. Perhaps it's just because so very few people outside the ranks of professional engineers know just what a magnetic amplifier is. Let's remedy that!

## Induced Knowledge

To start with, we need a pretty clear idea of what happens inside an inductor. A simple inductor has a winding which consists of insulated wire wound round a core of a soft magnetic material. Soft doesn't mean that you can spread it on your bread, but that the material magnetises easily, and demagnetises just as easily. Take a piece of this material, hold a magnet near it, and it's magnetised. Take a magnet away and it's demagnetised. This material we use for the cores of inductors, transformers, electric motors, relays etc

An inductor makes use of this 'soft' magnetism. The winding has an alternating current flowing in it. This alternating current (changing smoothly from a peak in one direction to a peak in the opposite direction and back) causes the core of the inductor to magnetise. The magnetism isn't steady like a bar magnet, but alternating, which is the point of using soft magnetic material. a graph of the magnetism (called flux density) of the core plotted against time would, ideally, have exactly the same shape as that of the waveform of the AC applied.

So far so good - it's an alternating magnet. But we've known for about 150 years (or someone has) that


Fig. 1. Control of a load using a variable inductor, this configuration has very little power lost as heat, unlike a resistive controller.
wherever there's an alternating magnetic field, any piece of wire or other metal will have an alternating voltage induced

Stick a piece of wire near your alternating magnet and you'll find an alternating voltage across the ends of the wire. The voltage is small if you use just a few centimetres of straight wire, but if you wrap several metres or wire round the core, so that all the magnetism of the core is at the centre of the coil of wire, then you find quite a respectable amount of $A C$. Recognise it?, a transformer.

## Laying Down the Laws

The laws of Electricity are very consistent, though, Any coil of wire around a core that has an alternating magnetic field will have an AC voltage induced. That means that if we have only one coil, and we send AC through to generate the magnetism, it will also have an $A C$ voltage induced in it. This voltage which the text books call a "back EMF", opposes the current which causes the magnetism which causes the voltage.

Result?
It's a darn sight more difficult to pass AC through an inductor than it is to pass DC!

When we use an inductor in a DC circuit, then apart from some effects at the moments of switch-on and switch-off the thing behaves like a resistance, good old Ohm's Law and all the rest, and a fairly low value of resistance at that.

Now you might think that it should pass the same amount of current for AC as for DC, but it doesn't.

Imagine that the resistance is 2 R , so that 10 V DC passes 5 A . Apply 10 V AC and the current's nothing like 5 A . It's not because Ohm's law stops working, it's because of the induced voltage. We're trying to push AC


Fig. 2. Simple magnetic amplifier circuit, showing DC control winding.
through with one voltage, and the induced voltage is opposing our efforts. It's only the difference between the two voltages that has any effect at all.

## Impedance Impediment

Suppose for example, that with 10 V AC applied, the induced voltage is 9 V 9 . This makes the difference equal to 0V1, and the current is

$$
\frac{0.1}{2}=0.05 \mathrm{~A}, \text { (by Ohm's Law) }
$$

Now these are calculations we seldom bother to make. Instead we measure a quantity called the selfinductance, $L$, of the coil and use this quantity and the resistance value to calculate impedance, which is the ratio

$$
\frac{A C \text { voltage }}{A C \text { current }}
$$

for the coil. In our example, 10 V causes 0.05 A to flow, making the impedence $10 / .05=200 \mathrm{R}$, not a particularly large impedance, but much greater than the resistance of 2 R .

The useful thing about an impedance is that there's practically no loss of power in it. Pass a current through a 200 R resistor, and you lose energy in the form of heat the amount of heat lost per second is $200 \times(\text { current })^{2}$ joules for a 200 R resistor. The same current through the inductor in our example doesn't look anything like this only its resistance loses heat, and that's only $2 \times$ (current) ${ }^{2}$ joules, because the resistance is only $2 R$.

We can therefore use an inductor to control the flow of $A C$ in a circuit (see Fig. 1) with none of the power loss that a resistor would cause. Now if we could just have a variable inductor, we could be very neatly control the flow of current in that circuit. Of course, we could use an inductor with tapped turns and slide contacts, built like a potentiometer, and we make use of just such a device, the familiar Variac. It's possible though, to control the inductance of a winding with no mechanical movement at all, and what makes it possible is the effect called saturation.

## Control-A-Coil!

When we send a current, AC or DC, through a coil of wire which is wound round a magnetic core, we can't pass as much current as we like and expect the magnetism to keep pace. At some stage in the game the core saturates, which means that it's as magnetised as it's ever going to be, no matter how much current is used. Now when a core is saturated like this, a change of current doesn't cause a change in the magnetism, so there's no more induced voltage. In other words, the inductance is no more and the impedance is practically zero,

Let the AC flow to it's load through an inductor whose core we can cause to saturate. How? By passing DC through another winding, by making the core of material which saturates easily, and the making the core continuous with no air gap.

That's our recipe for a magnetic amplifier.

## Amps For Amps

Figure 2 shows a simple magnetic amplifier circuit The inductor L1 has a large inductance when the core is not saturated, because of that, its impedance is very large, enough to make the current in the circuit very small. Now let DC flow through the second winding L2, and the core saturates.

If we can keep the core saturated for the whole of the AC cycle, then the inductance of L1 is almost zero, and the full amount of $A C$ current flows through the load.

We don't of course, have to switch between saturation and no-saturation. We can adjust the control current so that the core saturates only on half of the AC cycle, or in peaks so that the average current through the lead is controlled.

## Self Satisfied

Even such a simple magnetic amplifier has a lot of advantages, such as low power dissipation and high power gain, but better results are possible by using what is called a self-saturating design. Self-saturation is a form of positive feedback, using some of the signal current to assist the DC control current. Fig 3 shows a half-wave self-saturating circuit. The rectifier D1 ensures that only one direction of current flows through the coil L1 and the rated load current will cause the core to be close to saturation. The DC control current in winding L2 need only be quite small to cause the core to saturate on peaks, so that less power is needed to control the load current, and power gain is much higher.

Only half cycles are passing into the load, however, so that a full wave version is more desirable.


Fig. 3. Half-wave control using self-saturation.


Fig. 4. Full-wave amplification with self saturation, by positive feedback.

A full-wave self-saturating magnetic amplifier is shown in Fig 4. Two sets of windings are used, each handling half of the wave, with rectifiers ensuring that
the AC wave is split into its two halves.
In all these circuits, an additional inductor is used in the $D C$ control line to prevent $A C$ appearing in the control circuit because of transformer action.

## Going Straight

DC amplification? Simple enough, just rectify the output of the magnetic amplifier - the self-saturating full wave type already has two rectifiers included in the circuit and only two more are needed. More sensitivity? Add another winding to pass DC bias current, and the sensitivity increases because the bias can be set so that the core is very close to saturation

Nothing could be that perfect, there has to be a snag somewhere, and response time is it for magnetic amplifiers. Being slow beasts a sudden change of control signal may not cause much change in the output current until several cycles of AC have passed through. Nevertheless for stabilising AC supplies, for control of large AC loads and for high power gains magnetic amplifiers are not so easily displaced by electronics. There's not much to go wrong, they can be built to order, and they can be repaired.

Ever tried to put a new junction into a thyristor? E

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# STACE DIMMER 

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The first types of dimmer used, of which there are still some examples in older theatres, was a variable resistarice type which used either a variable or switched power resistor in series with the load. With small loads a wire wound resistor or a carbon pile was used while larger loads used a tank of saline solution with a central
electrode which was raised or lowered in the liquid, effectively changing the resistarice. This type of dimming, while reasonably effective, dissipated a lot of power which made life uncomfortably hot for the operator; since to minimise mechanical linkages the dimmers themselves were often in the control room:



## Electronics

With the advent of electronics, life was a little bit easier. The use of phase controlled dimming using thyratrons and later SCRs and Triacs reduced the heat dissipation dramatically (if you'll excuse the pun) and also allows the control to be physically separate from the dimmer. Besides being easier for the operator performances were greatly enhanced by the much better control available.

Today the use of phase control is almost universal as it is simple, reliable and cheap. Another method in use today is by magnetics; this type has the advantage of generating no RFI but unfortunately is expensive.

The problem of RFI is common to all phase control circuits, but can usually be reduced to acceptable, levels by the use of a choke and several capacitors. For RFI the choke need not be very large, but one other effect of phase control is the audible rattling of the lamp filament (especially with the larger globes) which is due to the sudden application of power, and the magnetic field so produced, each half cycle. This can be cured by reducing the rate of the rise of current by using a larger choke.

## Type Casting

We have given some schematic diagrams of types of dimmers which have been used previously. Fig. 1 is the oldest type comprising simply a variable resistor in series with the load. The second (Fig. 2), probably the most common type in use today. (mainly in homes) is very simple but lacks the versatility needed for theatrical work.

The third type (Fig. 3) is in common use and while still very simple does have many good features. These include having the

Fig 1. (Far left). The earliest type of dimmer employing just a variable resistor controlling the load.

Fig 2. (Left). Common! The most usual kind of light dimmer in use today.

control potentiometer isolated from the mains voltage and also a modified control curve to give a better input-output voltage relationship. Synchronization is referred to the zero crossing of the mains voltage, making the unit more suitable for driving inductive
(fluorescent) ioads; this also eliminates hysteresis which occurs with the simple dimmers.

The dimmer to be described here is more complex than most but a great deal of effort has been taken to ersure that a// problems have been solved. A low pass filter, with phase


Fig 4. The circuit of the control desk sections.


Fig 5. Power supply circuilt

## HOW IT WORKS ~CONTROLLER

There are two controls for each dimmer along with two master controls. The master controls vary the voltage on the individual level control potentiometers from 0 V (no light) to - 8 volts (full light). Normally one master will be at maximum and the second at zero. The outputs of the two controls for each dimmer are added by an operational amplifier, referred to 0 V . As one set of potentiometers has 0 V on both of its ends it
can be varred without changing the output allowing it to be set for the next scene. By varying the master controls together, but in opposite directions, the complete lighting set up can be smoothly varied from one scene to the next.

As we need +12 V out to drive the dimmers the supply voltage of the control desk is $\pm 15$ volts.

Fig 6. (Below). Showing the phase v frequency responses effect of compensation upon response


(SWy) $30 \forall 170 \wedge$ คกd」no

## PARTS LIST

| Resistors all $1 / 2 \mathbf{W} \mathbf{5} \%$ |  |
| :---: | :---: |
| R11,12,21,22,31 |  |
| 32,41,42,51,52, |  |
| 61,62,71,72,$81,82,91,92,101,102 ~$ |  |
|  |  |
| R13,23,33,43,53 |  |
| 63,73,83,93,103 | 220k |
| R14,24,34,44,54, |  |
| 64,74,84,94,104 IOOR |  |
| POTENTIOMETERS |  |
| 22 off 25k | 25 k 1 in .60 mm slide |
| RV3 | 5 k trimmer |
| CAPACITORS |  |
| C1, 2 | 33p ceramic |
| C3,4 | 220u 50V |
| C5,6 | 10 L 25 V |
| C11,21,31,41 | 33 p ceramic |
| C51,61,71,81 | 33p ceramic |
| C91,101 | 33p ceramic |
| SEMICONDUCTORS |  |
| IC1,2 | 301A |
| IC3 | 7815 |
| IC4 | 7915 |
| IC11,21,31,41 | 301A |
| 51,61,71,81 |  |
| 91,101 |  |
| Q1,2 | BD 140 |
| D1-D4 | 1 N4001 |
| MISCELLANEOUS |  |
| Transformer | 30 V |
| Box and front panel | 5W |
| Knobs to suit |  |

Fig 7. (Right): Component Overlay for the Controller Module.

## BUYLINES

> Apart from the pulse transformer T1 for details of which see Table One none of the components in this (admittedly huge) project should tax your local supplier overmuch. If you send us an SAE we will send you the foil patterns for the PCBs used here, as they were simply too big to print full size.

> Any 400 V ten or twenty amp triac will probably serve if you can't find the specified type easily.

correction, is used to ensure accurate synchronization. The control curve is also modified to give a subjectively more linear response and it has the ability to drive a fluorescent load without requiring a ballast resistor. Both the maximum and minimum light levels are adjustable without interaction giving reliable and predictable output. This is especially necessary if a dimmer fails for some reason and is replaced by a spare unit.

## The Protection racket

The protection of SCRs and Triacs, especially Triacs, is usually difficult as they tend to fuse faster than the fuse purportedly protecting them. The use of a cheap Triac which requires an expensive fuse to protect it is false economy. We have used a large rugged Triac (40 A device for the 20 A dimmer) which allows economical fuses to be used, especially for the 10 A version.

On the control side we will be describing a panel with two sets of long sliders per dimmer with two master controls which allow the next scene to be set up then faded in when required. A digital memory which can 'prerecord' scenes and recall them on demand may be published later.

## Dimmer Module - Construction

Assemble the boards with the aid of the overlay. The heatsink should be drilled and tapped for the triac to allow easy replacement if ever necessary. Note that the mounting of the fuse is different for the 10 and 20 A dimmers.

The choke is bolted onto the PCB using the long clamping bolts, preferably using rubber, grommets in the holes in the board (they may have to be drilled out to do this). The leads from the choke should be bent such that they go into the holes provided without going near the mounting bolts which are at earth potential. The leads can now be soldered (both sides on the 20 A unit).

The pulse transformer can now be added according to Table 1 Be careful when winding this transformer not to damage the insulation on the wire as there is 240 V between windings. We also recommend some epoxy between the transformer and the board.

The printed circuit boards for the two versions of the dimmer board are identical in layout and differ only in that the connector end of the 20 A board is double sided to present a greater area of contact with the connectors.

## Controller-Construction?

The component numbering system used on the controller drawings is designed to indicate which channel a particular component is part of. The printed circuit board drawing for the dimmer board is too large to publish in the magazine at full size; however, the pattern is available from our offices for the cost of an SAE - a large SAE!

If the dimmer modules are not required to be connected through sockets, the total cost can be reduced by connecting directly to the modules and mounting them in a box. In the 20 A unit the heavy wires should be bolted on to the appropriate pads to ensure contact to both sides of the board

One more modification to the control desk is the addition of a black-out switch which allows all lights to be blacked out without moving the master control. This is simply done by switching the supply voltage on the master potentiometers from the 8 V supply as set by RV3 to OV. RV3 should be adjusted such that with one master at maximum, the second at minimum and one
individual control at maximum that its output voltage should be +10 volts.

## Setting up

With the dimmer module the trim potentiometer has to be adjusted so that the output pulse from IC7 occurs at the very end of each half cycle. This is easiest set using an oscilloscope although an approximate setting can be made without one.

If the dimmer is connected up to a reasonably heavy load and adjusted for about $1 / 3$ level it will probably be found that with RV3 at one end the light level is not stable and tends to flash. This is caused by the sync pulse occurring after the end of the half cycle and the trigger pulses from the previous half cycle triggering the next. The trim potentiometer RV3 should be turned back about $1 / 4$ turn from the position at which this effect stops.

## Max and Min

When adjusting the maximum and minimum levels the minimum should be adjusted first. Note that the control potentiometer must be slightly up off zero to get any light and minimum should be adjusted at this point. The maximum should be adjusted with both the master and individual control at maximum and set to the point where the light level is just starting to drop.


Shown above is a completed dimmer module
PROJECT: Dimmer


|  |  |
| :---: | :---: |
| ~DIMMER MODULE |  |
| To help explain the operation the circuit can be broken into seven sections. | megohm input impedance and secondly it detects when the input voltage falls below 0.1 volt and turns the dimmer output completely |
|  |  |
| This is a simple full wave rectifier which be turned up to give a better control range, ie |  |
| gives about $\pm 18 \mathrm{~V}$ aft | nts just glowing, yet have |
| and C17. Using 3 terminal regulators this is them off if the control voltage is reduced to |  |
|  |  |
| circuitry. | If the voltage is above 0.1 volt, the diode D1 ill lift the voltage on pins of ICl to equal |
| 2. Control tone filter and sync generator | hat of the input on pin 3. However if the |
| As the name implies this removes the control | voltage falls below this level, the voltage on |
| tones that the supply authority superimposes | pin 2 will remain at about 0.1 volt due to R3 |
| on the mains voltage. These are normally and the output of ICl will go to about -10 |  |
| about 1050 Hz and can cause problems by |  |
| upsetting synchronization of dimmers. The |  |
| filter is a low pass type comprising IC6 and | 5. Mixer-compar |
| associated components. As filters always | IC2 mixes the input voltage, the output of the |
| alter the phase relationship this is corrected | curve generator the sync pulse and the |
| using phase shift networks. Cl1/R21 and | minimum adjustment potentiometers. This |
| C12/R22. Potentiometer RV3 is used to ens-ure the phase shift is zero (at 50 Hz ) with | gives the waveform shown in Fig. 2 with the |
|  | input voltage and the minimum adjustment |
| normal component variations. If the output of IC6 is between +0.6 volts and -0.6 volts, neither D5 nor D6 will be forward biased | only moving the curve up and down without |
|  | altering the shape. When the output of IC2 |
|  | falls below zero volts the output of IC3 goes |
| sufficiently to change the input voltages to | from -10 V to +10 volt with D3 and R8/9 |
| IC7'so its output will be -10 volts. As the | providing about 1 volt of positive feedback. |
| output voltage of IC6 is a 'clean' 50 Hz sine | The voltage has to rise to above IV to force |
| wave of about 6 volts amplitude this will only | the output back to -10 volts. The diode is |
| occur at a small region about the zero cros- | necessary to ensure that the voltage at the |
| sing point. At all other times the output of | input of the oscillator IC4 remains within the |
| 1 C 7 will be +10 volts. The result is a negative pulse, about $250 \mu$ s wide at the zero crossing |  |
|  |  |
| point of the 50 Hz . | 6. Oscillator/triac drive |
| 3. Curve generator | A CMOS oscillator IC4 is used to drive Ql |
| This produces the output shown in Fig. 6. | which supplies the energy for the pulse |
| When the sync pulse occurs, transistors Q2 and Q3 discharge capacitors C8 and C10. | transformer Tl. The oscillator will only |
|  | operate when the control inputs (pins 1 and |
| Immediately on release of the sync pulse the | $13)$ are +10 V . The frequency is controlled by |
| output of IC5 begins to ramp up slowly due to | C5 and is set at about 150 kHz . Resistor R13 |
| R16 charging C10. However, while initially | provides current limiting for the pulse trans- |
| the voltage across R14 is zero and therefore does not affect the charging of C 10 , as C 8 | former while R12 prevents the reverse vol |
|  | tage damaging Q1 if the load on the secon- |
| begins to charge due to R15 itse effect be- dary load (the triac) becomes dion,onnected. |  |
| comes more and more dramatic. A curve is |  |
| necessary as it gives a better input/output | 7. Power stage |
| voltage relationship but the curve must be reproduceable hence the circuit used. | This is simply a triac with a choke in series to |
|  | , prevent both RF1 and 'filament rattle' and a |
|  | fuse to protect against short circuits. |
| 4. Input buffer | Capacitors are also used as bypasses to help |
| This serves two purposes; firstly, it allows a | prev |


Fig 11. An expanded view of the
drive waveform showing 01
collector voltage.


[^2]

PARTS LIST

| RESISTORS all $1 / 4 \mathrm{~W} 5 \%$ |  |
| :--- | ---: |
|  |  |
| R1, $2,8,14,15$, | 100 k |
| $17,19,20,23-26$, | 4 M 7 |
| R3,27,30 | 22 k |
| R4 | 47 k |
| R5,6 | 10 k |
| R7,9,10 | 2 k 2 |
| R11 | 1 k |
| R12 | 470 IW |
| R13 | 2 M 2 |
| R16 | 15 k |
| R18 | 4 k 7 |
| R21 | 220 k |
| R22 | 3 k 3 |
| R28,29 | 10 R |
| R31 | 47 R 1 W |
| R32 |  |

POTENTIOMETERS

## RV1

1 M linear
RV2 100k linear
RV3 25k trimmer

CAPACITORS
C1,8,10,11,20
100n polyester
C2,4,9,15
C3, 7
C5
C6
C 12,13
C14
C16,17
C18.19
C21,22,24
C23,25
SEMICONDUCTORS
IC1-3,5,6,7
IC4
IC8
IC9
Q1
Q2, 3
TRIAC
D1-D6,11,12
D7-D10
33 p ceramic
10 n polyester
330p ceramic
100p ceramic
22 n polyester
$33 n$ polyester
220u 25 V
10 u 25 V
-1n polyester
$33 n 250 \mathrm{~V}$ AC

LM 301A
4011B
7812
7912
2 N 3643
BC 559
BTW41/400
1N914
1N 4001
MISCELLANEOUS
T1 see text, T2 $24 \mathrm{~V}, 5 \mathrm{VA}$, heatsink and choke and fuse 10A or 20A to suit, fuse holders

## TABLE ONE

The pulse transformer $T 1$ is the most difficult component in the project to find or produce. Tandy market a $4: 1$ device and this must be first choice. If this, and all other, commercial units prove elusive - try winding it yourself onto an ferrite ring of about 2 in outside diameter, using 50 turns and 12 turns for the windings to obtain the required ratio. Some experimentation may be needed here in order to get the triac to fire properly, and we do not recommend you try this unless you have wound coils previously.

## From Science of Cambridge: the new MK 14. Simplest, most advanced, most flexible microcomputer-in kit form.



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The MK 14 is a complete micracomputer with a keyboard, a display, $8 \times 512$-byte preprogrammed PROMs, and a 256 -byte RAM programmable through the keyboard.

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Yet in kit form, the MK 14 costs only $\ell 39.95$ $(+£ 3.20 \mathrm{VAT}$, and $\mathrm{p} \& \mathrm{p})$.
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2. 16-line RAM I/O device allowed for on the $\mathrm{PCB}^{\mathrm{P}}$ giving further 128 bytes of RAM.
3. Low-cost cassette interface module-which means you can use ordinary tape cassettes/ recorder for storage of data and programs.
4. Revised monitor, to get the most from the cassette interface module. It consists of 2 replacement PROMs, pre-programmed with sub-routines for the interface, offset calculations and single step, and singleoperation data entry.
5. PROM programmer and blank PROMs to set up your own pre-programmed dedicated applications.
All are available now to owners of MK 14.

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 As a computer, it handles operations of all types-from complex games to digital alarm clock functioning, from basic maths to a pulse delay chain. Programs are in the Manual, together with instructions for creating your own genuinely valuable programs. And, of course, it's a superb education and training aidproviding an ideal introduction to computer technology.
## SPECIFICATIONS

- Hexadecimal keyboard $\bullet 8$-digit, 7 -segment LED display $\bullet \times 512$ PROM, containing monitor program and interface instructions - 256 bytes of RAM - 4 MHz crystal -5 V regulator $\bullet$ Single 8 V power supply $\bullet$ Space available for extra 256 -byte RAM and 16 port I/O Edge connector access to all data lines and I/O ports


## Free Manual

Every MK 14 kit includes a Manual which deals with procedures from soldering techniques to interfacing with complex external equipment. It includes 20 sample programs including math routines (square root, etc), digital alarm clock, single-step, music box, mastermind and moon landing games, self-replication, general purpose sequencing, etc.

## Designed for fast easy assembly

The MK 14 can be assembled by anyone with a fine-tip soldering iron and a few hours' spare time, using the illustrated step-by-step instructions provided.

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## Gertroniostotey

## What to look for in the April issue: On sale March 2nd

## Amp Survey

Build-it-yourself hi-fi continues to flourish, and new designs appear almost daily. Power amplifiers are a favourite in the field, and their numbers, by now, are legion.

Unfortunately there is no way for the home constructor to 'listen in' to a module before he builds it, and thus he is left to fall back on the spec. sheets. Fine if you like it, rotten if you don't.

Next month we're surveying the field, giving full details of all the models we can find, and putting the market leaders against top quality commercial equipment to find out how they sound.

## MAINS SEEKER

So you are about to drill the living room wall to hang up those shelves you promised the wife 7 years ago. Black \& Decker in hand you advance to the plaster. Wait a minute there a mains socket right beneath.

Doubt sets in - to drill or not to drill - that is the question. Which way do the wires run? Will you black out the entire Universe if you try it? How can you find those wires?

Simple really - just read ETI next month when we have a neat little project to show you exactly where the mains wires lie!


## AMBUSH



## OCTAVE SHIFTER

A superb little circuit to add that instant 'jump' to guitar playing. Operated by a footswitch the effect has a unique sound all its own - not to be missed - no strings attached to this one.

## 3080

Well ten of them anyway The 3080 is a much under-rated device, and next month's IETs circuit man Tim Orr hopes to put that right with ten ways to use device, all comprehensively explained to help you design the other 3070 circuits yourself.

Your starship crashes through the void - running between the lines of enemy dreadnoughts to deliver medical supplies to the seiged plant of Tora. In order to preserve energy your ship has no weapons, only its shields and its speed.

Missiles can appear from any direction, and to destroy them you must actuate your shields at the precise moment of impact, thereby conserving power and allowing the engines to keep you moving at Warp Factor 20.

Can you make it through the Ambush and make Capt. Kirk look a cissy?

# microfile 


#### Abstract

ALTHOUGH out of the first flush of youth I still consider myself to be young at heart so the young computing funfair seemed just the thing for me. Organised by the British Computer Society (BCS) the fair filled the Bloomsbury Centre Hotel early in January with the younger members of the ever growing group of people with an interest in computers.


Following on from the 'Living with Computers' conference, again organised by the BCS and held in the Institute of Education near the Bloomsbury Centre, the event provided a fitting climax to a very stimulating few days.

The funfair included exhibits ranging from minicomputers, computers being used for choreography in ballet, computer controiled games, DIY computer kits, a micro-computer controlled railway and many stands showing how the computer is in use in our everday life. Among this last group of exhibits were the Abbey National Building Society (they got the computing habit, they have) and the police, showing that big brother was alive and well at the funfair

The NASCOM stand was a great attraction for many people, demonstrating as it did the latest add ons to that popular DIY computer kit, the NASCOM 1. The buffer board, mother board and expansion RAM cards allowed the NASCOM's on display to run the 2 K BASIC interpreter NASCOM are now producing. The BASIC in cludes all the facilities common to basics of this size, interger arithmetic, 26 variables (Designated A-Z), single dimension array plus assorted commands, operators and functions. In addition a machine code call greatly extends the power of the interpreter. NASCOM's super tiny BASIC makes use of this machine code call to provide amongst other commands, an edit function which allows the insertion/deletion of individual characters within a line, a renumber command and a facility for string inputs - something sadly lacking in the basic BASICS.

Undoubtedly though the exhibits that caused the greatest interest amongst the younger generation that made up the majority of the visitors to the fair were the rows of amusement machines. Everything from cowboys at the OK coral to star wars in space quadrant 0040. 7689. Microprocessors have revolutionised the arcade industry, both in bringing arcade type games into the home and in dramatically increasing the sophistication and supposedly, entertainment value of the machines in the public domain. Certainly the young audience were impressed.

Altergo ran a painting competition in conjunction with the funfair and Saturday saw the prize giving ceremony. The subject for the painting was "My Friendly Computer." The first prize was an Altergon robot, a beast that walks, talks, moves and flashes its eyes. Talking of beasts brings us onto beauty and Joanna Lumley - of the new Averigers - who presented the prize to the winner of the competition - 12 year old Fiona Mackay.

As well as the exhibition stands a concurrent series of lectures was presented in a hall adjacent to the main event. The lectures concentrated on introducing people to the various aspects of computing and the careers potential offered by the computing industry. These were well attended in spite of the attractions of the afforementioned amusement machines.

All in all a successful gathering that introduced the fascinating and diverse world of the computer to people who will form the systems engineers, computer operators, engineers etc. of tomorrow.


Any excuse to get a robot onto the pages of ETI is eagerly taken up. The fact that there are a couple of ladies in frame is incidental. The robot was first prize in the painting competition organised by Altergo. Fiona Mackay won the robot which was presented by Joanna Lum/ey.

Following on from my item last month concerning low cost keyboard designs Mr Charles Lacey has written to me with details of a project along these lines that is at the prototype stage. Designed as a touch keyboard with fifty keys including space bar, 2 shift keys, a delete key and two spare controls the system is more elegant than my attempt. Looking back at the circuit I proposed it does seem a bit clumsy

With luck Mr Lacey should have a very nice project in a couple of months time. We'll keep in touch.

By the way if anybody else has had ideas along these lines please let me know.

Finally may I add my own tribute to John MillerKirkpatrick who died last December. I'd known John just on two years and it was the System 68 project that brought us together. John handled the design and construction of the system while I dealt with the production from the magazine's end. System 68 had many teething problems but it was the first such project tackled in this country, way ahead of its time, and John remained enthusiastic throughout the problems and put in vast amounts of time to get things sorted out.

The last time I saw John was at the Breadboard exhibition. He took a keen interest in all of the stands and no doubt had a few ideas of his own for things to tackle in the future.

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$$
\mathrm{P}=\mathrm{E} 1
$$

In a DC circuit where both voltage
and current remain constant no problem arises. However in an AC or a DC circuit where the voltage is not constant with time, this formula only holds for instantaneous power as the power varies with time. Power as we usually use the term is the time average of this. If the load is resistive, i.e. contains no inductance. or capacitance, and we can measure the RMS value of the voltage, we can still use this simple formula. However measuring the RMS voltage is not easy as most voltmeters measure the peak or average rectified voltage with a s'itable scaling factor built in to give a correct result when measuring a sine wave signal.

## Reactive Reaction

If the load is reactive the current and voltage will no longer be in phase,
i.e. the peaks do not occur at the same point in time. The difference can be expressed either by the phase angle in degrees or by the cosine of this angle known as the power factor). The current waveform can either be ahead of the voltage (leading) or behind it (lagging). Capacitive circuits give rise to a leading power factor while inductive circuits lag.

If working with a sine wave, and if the power factor is known, the formula for power can be expressed as:

$$
P=E \mid \cos \varnothing
$$

where $\varnothing$ is the phase angle. In a DC circuit $\cos \varnothing$ is unity so the formula holds for this case as well. An example is a 40 W fluorescent light which takes 430 mA from the 240 V mains. At first sight, this implies a

power consumption of over 100 W , until it is realised that its power factor is about 0.45 lagging. The formula above, using $\cos \varnothing=0.45$, thus gives a power consumption of only 46.4 $W$. (The additional 6 odd watts is dissipated in the ballast.) The product of voltage and current is known as the VA rating and is used when calculating the currents in a circuit. If a capacitor is connected across a sine wave $A C$ circuit the current taken can be calculated by dividing the voltage by the reactance of the capaitor. While this circuit draws current, it has a power factor of very near zero ( $90^{\circ}$ phase lead) and therefore takes no power! By adding the correct amount of capacitance to an inductive circuit (i.e. the fluorescent light) the power factor can be altered, reducing the current drawn (but not the power).

Confused yet?

## Ample Reason

Getting back to audio amplifiers and their ratings, the problem lies in the complex nature of the music waveform and how to specify the amplifier's rating. As the waveform is far from a constant sine wave with the peak power being anything up to 20 times the average, numerous methods such as peak power, peak to peak power, music power, etc. evolved. However, for a long time there was no set standard, and one amplifier advertised with a 50 W (music) rating was in fact a 5 W stereo amplifier. The situation got so out of hand that the US Government brought down legislation on how amplifiers were to be tested. This is with a continuous sine wave signal with level set so that the distortion is at a specified level and power calculated from the RMS output voltage: hence the term RMS power. Note however that the term RMS refers to the method of measurement, i.e. the use of RMS voltage, and it is not the RMS value of the power waveform. It is, in fact, the average of the power waveform.

Speakers are just as confusing. They are normally specified not in terms of the power they can dissipate, but the maximum power of amplifier they are suitable for. This is due to the fact that music is never (well, rarely) a continuous sine wave and the average power in the speaker may be only $10 \%$ of the RMS rating of the amplifier, even with the amplifier clipping.

To measure the power actually being delivered to the speaker under music conditions, a wattmeter must be used

## Design Features

To multiply current and voltage together we had the choice of . analogue or digital techniques. Unfortunately while digital is the 'in' thing, offering versatility and accuracy, it is not fast enough to calculate the instantaneous power on high frequencies. We therefore chose the analogue method.

Looking around the ICs, the only ones with reasonable price and availability were the MC 1494,1495 and 1496. The 1496 (or 796) is the cheapest and most readily available, but has the disadvantage of not being able to multiply DC signals or AC signals with a DC offset. The 1494 and 1495 are about the same price, and of the two, the 1494 was more linear and easier to use.

We chose not to use any input buffer on the voltage input but had to pay the penalty of having a lower input impedance than normal with voltmeters.

## Using the Power Meter

To use the meter we must measure both voltage and current. There must be a common point for these measurements. The current
connection can be in either of two ways as shown in the drawings below. One measures the power out of the supply and the second the power into the load. The difference? The current shunt in the wattmeter drops one volt when working at the full range value and this may or may not affect the reading. At 10 A this accounts for 10 W which, if the power being measured is only 100 W , is a $10 \%$ error - although if the measured power is 2400 W the error is only $0.4 \%$.

The range of the meter is the product of the individual ranges, i.e. on 30 V and 1 A the fsd is 30 W , while 30 V and 3 A gives 100 W FSD. To help give a reading reasonably high on the scale, the voltage range can be overvoltaged by a factor of 2 . Due to power dissipation problems this should not be attempted on the current ranges The peak voltage or current can be as high as three times the range value.

## Construction

We mounted all the components associated with the meter and the switches on a single pc board and if the same or similar case is to be used this is recommended.

Except for the meter and the switches the components are mounted on the 'normal' side of the pc board. These should be mounted



## HOW IT WORKS

Power is the product of current and voltage. This holds irrespective of the nature of the load, provided you are talking about instantaneous power. By multiplying current and voltage together and then taking the average of these instantaneous values we find the true power. Again this works irrespective of the load.
In this circuit the multiplying is done by ICl (MC1494), the output of which is a current proportional to the product of the inputs. For more detailed notes on this IC, see the separate section. The current output of this IC is converted to a voltage by IC2 with C2 providing the averaging. The meter is then simply wired across the output of this IC with a meter reversing switch provided. This reversing switch is needed not to measure negative power, but to correct for reversed readings due to differing external connections.
The power supply is a full wave bridge with a centre tap giving about $\mp 20 \mathrm{~V}$ DC which is then regulated to the $\mp 15 \mathrm{~V}$ required by ICl.

Adjustments for zeroing the voltage and current inputs are provided by RV2 and RV3 while RV1 compensates for offsets in the output, These are supplied by a stable $\mp 4 \mathrm{~V}$ reference in ICl. Range switching is done by SW1 and SW2. Protection against overvoltaging the IC is provided by Dl D4.

Fig. 1. The circuit diagram of the audio power meter.


Fig. 2. Power Supply Circuit.

Right: meter scale designed for a 1 mA FSD meter. These scales may need to be altered for differing meter units.


## PARTS LIST



Fig. 3. Overlay for the Power Meter.

| Resistors all |  |
| :---: | :---: |
| R1, 7, 24, 26 | 1k |
| R2 | 100k |
| R3 | 470k |
| R4, 20 | 15k |
| R5 | 220R |
| R6 | 3k3 |
| R8 | 330R |
| R9, 19, 21 | 10k |
| R10 | 1R1W |
| R11-13 | OR22 5W |
| R14 | OR685W |
| R15 | 2R2 1W |
| R16 | 6 R 81 W |
| R17 | 22R |
| R18 | 47k |
| R22 | 18k |
| R23, 25 | 6k8 |
| R27, 28 | 330k |
| R29 | 4k7 |
| POTENTIOMETERS |  |
| RV1-3 | 20k trimmer |
| RV4 | 5 k trimmer |

CAPACITORS

| C1 | 33 p 500 V ceramic |
| :--- | :--- |
| C2 | 330 n polyester |
| C3 | $33 p$ ceramic |
| C4, 5 | 220 u 35 V electrolytic |
| C6, 7 | 10 u 25 V electrolytic |
| C8, 9 | 100 n polyester |

SEMICONDUCTORS

| IC1 | MC1494 |
| :--- | :--- |
| IC2 | $301 A$ |
| IC3 | 7815 |
| IC4 | 7915 |
| D1-D4 | 1 N914 |
| D5-D8 | 1N4004 |

MISCELLANEOUS
PCB
SW1. 2 two pole 6 position 10A rotary
Radiospares
SW3, 4 two pole toggle switches
Transformer 15-0-15, 5VA
Meter 1 mA FSD
Three binding posts
Instrument case $255 \times 100 \times 205 \mathrm{~mm}$
Power cord and clamp
Two knobs
Front panel

## BUYLINES

Most of the parts for this project are readily available. Two things which may cause trouble are the switch assemblies and the quadrant multiplier itself.

The switch is an RS unit and as such can be obtained from any of their stockists. As for the IC. Tamtronik - who advertise on page 32 of this issue - can supply this and by the time you read this they will be able to sell you all the rest as well!
first with the only critical part of the assembly in the area of the range switches. Here the high powered resistors should be spaced at least 5 mm from the PCB as they run hot at maximum current. Also the leads of all the reistors in this area should be cut off close to the pc board after soldering. This is to give adequate clearance to the rotary switches. We used two self tapping screws into the plastic of the transformer case to help fix it onto the board. We have made


Fig. 4. This connection measures the power into the load.


Fig. 5. This connection measures the power out of the supply.


Fig. 6. Transfer characteristics of the IC.


Fig. 7. Typical connections for a wide band multiplier or balanced modulator.


Fig. 8. The internal circuit diagram of the IC.


Fig. 9. Typical connection of a low frequency multiplier, For a squaring circuit simply parallel the two inputs. In this case pin 6 can be connected to OV and P1 deleted.


Fig. 10. Typical connection of a divide circuit. For the square root joins pin 9 and 10 . Like the squaring circuits pin-6 can be connected to OV and P 1 deleted.

Full size foil pattern for the power meter.


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# audiophile VISUALY HI-Fl? sans mesemenmp 



BY NOW I expect most of you have already heard of the TA-E88 pre-amp, the Sony flagship design. Costing a mere $£ 699$ it has been designed to match the TA-N88 VFET power amp, and uses FETS in the later stages of its circuitry. Overall the finish of the unit is probably up to a $£ 700$ standard. All sockets are gold plated, and a gold ended twin phono lead is supplied as standard. The controls are very nice to operate, and the volume and balance controls are very special indeed. Stepped attenuators are employed, but the operation is so smooth as to make you doubt it.

As you can see from the internal shot below, the signal path lies entirely along the PCB. There are no leads from the board carrying signal potentiais, all switches and sockets are mounted in place, and extended to the front panel where need be.

## Shifting Load

There is even provision for switching about the resistance and capacitance associated with the magnetic phono inputs. Gold plated switches of course. Adjustment is variable between 10k-100pf and 100k-500pf. A useful provision this.

Completely separate channels - and PSUs - keep the right insensitive to the meanderings of the left and with the moving coil inputs especially this can be no bad thing.

The head amp incorporated is a version of Sony marketed HA55, one of the best mains powered units on the market, and has two possible input impedances.

All this and no tone controls. It should be interesting to go through the circuit section by section, so I suppose the place to begin is the beginning.

*


## Head Start

As shown below, the moving coil amp consists of a differential pair with current minor driving an emitter follower output stage.

The differential circuit consists of cascade connected transistors to get the noise and gain figures required. and 44 dB of negative feedback is applied to lower distortion as usual. Low impedance feedback paths like this are fine for some applications, but need careful design indeed to avoid becoming more of a hindrance than a help.

used in both the second differential amplifier and in the constant current stages of the circuit.

The output is once more an emitter follower following an emitter follower, and the components used are the expensive metal film resistors and polypropylene capacitors. Still when you've got $£ 700$ to spend - why not eh?

Equalisation is unusually accurate at $0.2 \%$.


Magnetic cartridge input
board.


## Standing At The Buffers

Buffer stages are liberally employed in the TAE88, between source inputs (non phono) and selector, and then either side of the volume control. Sony call the output amp a 'flat' amplifier for no reason I can fathom.


## Buffer amplifier circuitry.

Above: moving coil input circuit and performance graph.

Left: the PCB from the balance control. Each 'step' is connected to a precision resistor!

## Magnetic Charm

For more normal pickups the TA88 employs dual FET inputs (in cascode mode) and more conventional RC feedback equalisation circitry. The FETs used were developed specially for the amplifier, when your 'Sony' of course you can get these things done.

The second stage is a differential amplifier also, to further stabilise and give the overall circuit greater immunity from current source drift. Duat transistors are


You can see the circuit for the buffer circuit opposite. Once again dual FET inputs, and if you thought that there is a very strong similarity between this and the phono amps, I don't think you'll get many arguments from me.

The 'flat' amplifier (below) differs in as much as it is designed to work into the load presented by cable and power amp. To do this without loss of frequency res-


Output buffer circuit.
ponse, a design closely akin to: a power amplifier configuration has been adopted.

Output impedance is about 100 ohms, so that fairly long interconnection runs can be tolerated, and up to 15 V can be safely output at around $.001 \%$ THD.

## Lugged Around

After reading through all the imposing technical info supplied with the unit I was almost afraid to wire up the box into my merely mortal system. I suppose I suspected some form of electron snobbery whereby the TAE88 would refuse to 'talk' to any power amp of less than immaculate pedigree.

In practice however it was a case of 'noblesse oblige' and the Sony worked impeccably with the rest of the universe.' Several power amplifiers were tried, including the Lecson AP3II and a Crimson set up.

At first the TAE88 sounds very impressive with a particularly good bass end. The treble is a little thin, but nothing to comment adversely on. After a while though I came to suspect that maybe the unit wasn't as good as I thought at first, and perhaps adds a certain metallic quality to the sound

Using the unit is a treat of the first order, and it inspires confidence better than Mr Callaghan ever did. Reservations must inevitably include that optimistic price level, and the less than perfect (just!) sound quality - which is as close to excellent as any other (but no closer!), but is more expensive approximation

A lovely machine nontheless and if Sony can pull down the price (exit gold?) one which would have received a wholehearted recommendation. ETII

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## J. Master.

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IC1,2 are RS Voltage to Frequency Chip
No. $307-070$
RV1,2,3 are 50k 20 turn trimmers

## Keyboard Guitar

A. Parker

The purpose of this project is to convert the waveform from a guitar or other instrument into pure square or pulse waveforms of the same frequency. The circuit is basically a frequency to voltage converter feeding a linear VCO.

The construction is straightforward provided the usual care is taken with the Cmos chips. For RV1, 2 and 3 we suggest 20 turn presets
as these will be needed for fine tuning of the circuit later. Also as an aid in testing we suggest that VR1 should NOT be soldiered in until after initial testing has been completed.

The tuning of the circuit is best done using a Meter, PSU, Signal Generator and frequency meter if possible. First set the sig gen to some suitable frequency (ie 100 Hz ) and using the meter between point A and earth adjust RV2 to give a voltage according go the formula

$$
V=F_{\text {in }} \times 10^{-3}
$$

(for $100 \mathrm{HzV}=100-\mathrm{V}$
Now using an ACCURATE PSU set point B to +1 V and using VR3 adjust the output to 1 kHz then set to +10 V and adjust to 10 kHz . Now solder RV1 and adjust until

$$
F_{\text {in }}=F \text { out }
$$

(NB This is a gross over simplifcation and patience is vital. Remember the price of the Chips before you throw them out of the window).

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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
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| 7425 | 20p | 7496 | 45p | 74165 | 60p | 4018 | 55p | $\text { CA } 304$ | $\begin{aligned} & \text { Op } \\ & \mathbf{6 0 p} \end{aligned}$ | LM 555 | 25p | SN 76228 N | 180p | TDA 2020 | 300p |
| 7426 | 22p | 7497 | 120p | 74166 | 75p | 4019 | 40p | CA 3060 | 225p | LM 709 C | 40p | SN 76660 N | 75p | $\text { TL } 084$ | 120p |
| 7427 7428 | 22p | 74100 | 80p | 74167 | 160 p | 4020 | 50p | CA 3065 | 200p | LM 710 TO5 LM 710 DIL | 60p | TAA 300 | 100p | $\left\lvert\, \begin{aligned} & \text { XR } 320 \\ & \text { XR } 2206 \end{aligned}\right.$ | 250p |
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The circuit shown allows a player to play tennis or squash against a perfect opponent, which is useful if one wishes to practise and cannot find another player.

The circuit 'plays' tennis or squash simply following the ball up and down the screen, thus it is always in the right place in order to hit the ball.

## B. Harvey

Although the circuit appears simple, (it only uses one gate from one IC!) the way it works is quite complex, suffice to say that it relies upon the way the AY-3-8500 games chip determines bat position from the setting of the hand controls.

The only modifications to the TV game are: (i) One lead connected to the ball video output of the games chip.
(ii) A switch wired in, selecting either a manual or an automatic player on the right hand bat.
(iii) This may not be necessary in home built games that use CMOS video mixers, but may have to be used in commercial units that sometimes use diode mixing circuits. The modification is shown and uses gates from the same IC. This will give a brighter bat and ball which is useful when playing squash.

## Sequence Switch

B. Willis.

The circuit right was designed to enable three relays to be individually switched by their appropriate buttons but such that only one relay can be energised at any one time. When any one relay has been energised the corresponding collector falls to near zero volts, which is connected to the base of the remaining two transistors; now if another relay is attempted to be energised the base of it's transistor will remain bottomed and keep the relay off. The rest button must be pressed before another relay can be energised. DI ensures that each transistor is kept off until the voltage applied to the base exceeds 0.6 V .
The flip-flops and push buttons can of course be replaced with standard switches if momentary action is not required

The circuit was used to control three radio transmitters where it was important that two should not be

switched on at the same time. The circuit lends itself to further applications; for example, switching various
inputs into an amplifier where it can replace the self-cancelling selector buttons.

## ambit tamatein

The PW Sandbanks Metal Locator: a kit based on this recently published design for this uniquely effective type of metal locator is available for only $£ 35.00+8 \%$ VAT The kit closely resembles the appearance as published, except that a close fitting injection molded housing replaces the vacuum molded electronics box - to improve the enviromental suitability of the construction. Carriage for complete kits $£ 1$. The New Catalogue - "Tecknowledgey Part 2
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ogc'd IF preamp
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limiting FM preamp Communications circuits
Communications circuits $\begin{array}{ll}\text { SD6000 } & \text { DMOS RF/Mixer pair } \\ \text { KB4412 } & \text { Bal mixers, IFtage }\end{array}$ KB4412 Bal mixers, I IFtage
KB4413 AM/SSB det. squelch, KB4417 mic processor MC3357 best thing in NBFM yet MC1496P popular double bal mixer 1

Multiplex decoders + noise blanker MC1310P \begin{tabular}{lll}
<br>
UA758 \& popular PLL decoder \& 2.20 <br>
\hline

 

<br>
\& buffered 1310 <br>
\& 2.20 <br>
\hline
\end{tabular} CA3090AQ RCA PLL decoder

HA1196 improved PLL HA1196 improved PLL decoder with stereo preamps
19 kHz pilot cancel, low distortion, high $\mathrm{S} / \mathrm{N}$ as HA11223 with remote vCO kill facility stereo MUTING preamp for post decoder mute
impulse noise blanker

Catalogue part 1:45p, part 250 p Ambit Items is generally $12 \% \%$, except where marked (*) £3. Phone Brentwood ( 0277 ) $216029 / 227050$ 9am-7pm. Callers welcome inc. Saturdays

## At lust. DIY HiFi whith loaks us if it isn't.

That's not to say it doesn't look like HiFi - just that it doesn't look like the usual sort of thing you have come to associate with DIY HiFi. The Mk3 outstrips and outperforms all British made HiFi tuners, and most imported ones too. Certainly at the price, there isn't one near it. But more than that, it looks superb. A small pic here would be an insult, so send an SAE for details on the kit that looks as if isn't It's something else..............

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* Exceptionally high performance - exceptionally straightfonward assembly - Baseboard and plug-in construction. Future circuit developments will readily plug in, to keep the Mkill at the forefront of technical achievement
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* Matching both the style and design concepts of the MkIII HiFi FM tuner
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With styling and dimensions to fit in with the rest of AMBIT's new range of tuner \& audio equipment

When the new range of OKI digital frequency display ICs was announced, the original prototype of the Dorchester had been made - but since so many of you wanted to use the OKI frequency counterdisplay system with the Dorchester, we quickly designed a unit to incorporate the necessary facilities. The Digital Dorchester is designed in 19 inch form, and forms a perfect match for the other units in the range. If you don't want to go to the expense of the full Ambit DFM1 module, with AM/FM/Time/Timers, then the MA1023 clock module can be used instead
The Dorchester has been described in PW Dec., Jan. and Feb. issues - but for those of you who may have missed it - it is an All Band broadcast tuner, covering LW/MW/SW and FM stereo in 6 switched ranges. Construction is very straightforward, with all the switching being PCB mounted and the revolutionary TDA1090 IC used for AM/FM.
The electronics for the radio section of the Dorchester remain unchanged at \(£ 33.00\), with \(12.5 \%\) VAT. The harciware package, of case, meter, PSU now costs \(£ 33.00+8 \%\) with \(12.5 \%\) VAT. The harciware package, of cas
with the MA1023 available for an extra \(£ 5\) only
For the fully digital version, with Ambit DFM1, the price is \(£ 56.50+8 \%\) VAT.

\section*{2 Gresham Raud, Brentwand, E5ser.}

\section*{AUDIO MODULES}

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\section*{RF MODULES}

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\section*{7 RF 030}

Improved version of above extra gain stage imposed \(S / N\) ratio and 1.5 \(\mu V\) sensitivity for 26 dB S/N way selector switch AFC stereo/mono switching - two additional inputs £19.95.

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Imput Impedance-1 Meg/40pt in shum
Imput Yothage-Max - 600 P P-P
honizomtal axis (x)
Deflection Sensitivity - 0-400nv/division Bandwidth [between 30B points) - \(1 \mathrm{~Hz}-350 \mathrm{KHz}\) Gain Conurol-Coninimous. when time base in EXI position Input Impedance 1 mos
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[^0]:    PORTWAY INDUSTRIAL ESTATE
    ANDOVER, HANTS SP10 3NM
    ANDOVER
    (STD 0264) 64455

[^1]:    Standard Features

    - Uses the ultra powerful 6502 microprocessor
    - 8K Microsoft BASIC-in-ROM
    - Full feature BASIC runs faster than currently available personal computers and all 8080-based business computesr
    - 4K static RAM on board expandable to 8 K
    - Full 53-key keyboard with upper-lower case and user programmability
    - Kansas City standard audio cassette interface for high reliability
    - Full machine code monitor and I/O utilities in ROM
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    ## Extras

    - Available expander board features 24 K static RAM (additional mini-floppy interface, port adapter for printer and modem and OSI 48 line expansion interface.
    - Assembler/editor and extended machine code monitor available.

[^2]:    Fig 9. Waveforms shown are:
    Sync pulse (output IC7), curve Sync pulse (output IC7), curve
    generator (output IC5), mixer output (output IC2), osciliator
    output (IC4), transformer drive

    ## (Q1).

