
(ii) ditionis 5


> Active Fliter stenal Tracer Casio FX502.

## TRANSCENDENT 2000 SINGLE BOARD SYNTHESIZER

LIVE PERFORMANCE SYNTHESIZER DESIGNED BY CONSULTANT TIM ORR (FORMERLY SYNTHESIZER DESIGNER FOR EMS LIMITED) AND FEATURED AS A
CONSTRUCTIONAL ARTICLE IN ELECTRONICS TODAY INTERNATIONAL.
The TRANSCENDENT 2000 is a $\mathbf{3}$ octave instrument transposable 2 octaves up or down giving an affective 7 octave range. There is portamento. pitch bending, a VCO with shape and pitch modulation. a VCF with both low and high pass outputs and a separate dynamic sweep control, a noise generator and an ADSR envelope shaper. There is also a slow oscillator, a new pitch The kit includes fully finished metalwork, fully assembled solid teak
The kit includes fully finished metalwork, fully assembled solid teak cabinet, filter sweep pedal, professional quality components (all complete - right down to the last ne $1 / 2 \%$ metal trim!) and it really is There is even 13A plug the last nut and bolt and last piece of wire, parts before plugging in and making great buy absolutely no more components are on the one professional quality fibreglass PCB printed with component locations. All the controls mount directly on the main board. all connections to the board are made with conneclor plugs and construction is so simple it can be built easily in a few evenings by almost anyone capable of neat soldering! When finished you will possess a synthesizer comparable in performance and quality with ready-built units selling for between $£ 500$ and $£ 700$ !

## COMPLETE KIT ONLY <br> £168.50 + VAT!

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COMPLETE KIT ONLY $£ 299.00$ + VAT!
To add interest to the sounds and make them more natural there is a chorus/ensemble unit which is a complex phasing system using CCD (charge coupled device) analogue delay lines. The overall effect of this is similar to that of several acoustic instruments playing the same piece of music. The ensemble circuitry can be switched in with either strong or mild effects
As the system is based on digital circuitry digital data can be easily taken to and from a computer (for storing and playing back accompaniments with or without pitcn or key change, computer composing etc., etc.) and an interface socket ( 25 way $D$ type) is provided for this purpose.
Although the DPX is an advanced design using a very large amount of circuitry. much of it very sophisticated, the kit is mechanically extremely simple with excellent access to all the circuit boards which interconnect with multiway connectors. just four of which are removed to separate the keyboard circuitry and the panel circuitry from the main circuiry in the cabinet
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# Quartz Melody Multi-Alarm Chrono For 1980 Try this 34 Function 

## Count-down Timer

Can be used for a host of applications from boiling an egg to warning you your parking meter is expired. The timer is presettable to $<3$ hours 59 mins. 00 secs. in 1 min . steps and counts down in 1 sec . steps. It operates quite independently of the other counters and the watch can be in any other mode whilst it is being used.
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## Alarm



The alarm can be set at 1 minute intervals to any time within the 24 hour period. A clear firm musical tone sounds for 1 minute at the appointed time. An automatic roll-over to the normal time is a feature after the alarm has been read. A clear indicator displays whether the alarm is set or not.

## Time Zone 

The time zone enables you to tell the time in two places at once. It can be useful on holiday or business trips. Just programme the second time zone and it will be permanently recorded for your easy reference.

## Chronograph <br>  18.NETU

This watch incorporates a sophisticated and very accurate stop/start counter which has many applications in sporting events and timing for recordings etc
Mode 1: Is the normal stop-watch mode. Stop Start-Zero.
Mode 2: The lap timer enables first and second past the post times to be recorded. The display is frozen but the counter continues to count
Mode 3: Longer timing intervals, such as journey times, can be recorded whilst the watch is reading its normal time. or the count-down is being used. The counter counts to 1 hour in $1 / 100 \mathrm{sec}$ steps in all its modes.


## MODEL M30

Display Format (NORMAL TIME DISPLAY)
 S2 Press S2.

5 independent working modes
i) Normal watch
ii) Count down alarm
iii) Alarm
iv Dual time zone
v) $1 / 100 \mathrm{sec}$. chronograph

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It has fast and slow setting rates for the counter and the alarm as well as the normal time setting.

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Features and Specification
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This design, published in Wireless World, although straightforward and relatively low cost provides a very high standard of performance. There are separate record and replay amplifiers and switchable equalisation together with a choice of bias levels are also provided. The
mechanism is the Goldring-Lenco CRV with electronic speed control.


COMPLETE KITS: Our complete kits really are complete. All of the projects shown on this page are supplied with tully finished metalwork, ready assembled high quality teak veneer cabinet (last 4 kits on this page), or professional quality rack mounting cabinet (first 2 kits on this page), cables, nuts, bolts, etc., and full instructions - in fact everythingi
All of the kits shown on this page are available as separate packs for those customers who wish to spread their purchase or perhaps make their own cabinets or metalwork. Prices are given in our FREE CATALOGUE

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use it to listen in on Radio 1 (or music, if you prefer) without driving the boss round the twist.
I would also recommend it whole-heartedly to insomniacs. While all about you are deep in dreamland, tune into your favourite middle-of-the-night music without starting a riot.
The TP-1E from JVC should be available for around $£ 4.60$ including VAT.


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With the Poly-Tester you can
test continuity, neutral and ground terminals in power outlets, semiconductors and voltages. There are no test leads to connect - the Poly-Tester has only one integral test probe, with a protective cover. It can also be used as an ordinary screwdriver-type voltage tester with the added advantage of being unaffected by induction. Connection to the user's body is by a metal pocket clip. Like the combi-Sensor, test indication is by sound and light.

The Combi-Sensor and PolyTester are $£ 7.60$ each plus VAT from C. Lord, 61 Forknell Avenue, Wyken, Coventry CV2 3EN.

## Opticom

A 96 kms experimental optical communications system has been set up by Philips Research Laboratories in Holland. It is believed to be the longest in the world.

The glass fibre connection consists of 16 reels of cable, each made of six separate strands. The reels are connected in two sections of eight kilometres each.

The signal is passed down one strand in the cable, through a repeater at each eight kilometre section, through the second eight kilometre section, then through a further repeater before returning up the cable via a second strand. This sequence is repeated via all six strands and eleven repeaters until the signal emerges at the
end of its 96 kms journey Attenuation is a mere $4 \mathrm{~dB} / \mathrm{km}$, including the juints every kilometre. Hence the need for repeaters at only eight kilometre intervals (conventional copper cable links need repeaters every two to four kilometres).
Japan isn't far behind the Dutch achievement. The Nippon Telegraph and Telephone Corporation have succeeded in transmitting an $800 \mathrm{Mbit} / \mathrm{S}$ signal through 30 kms of fibre-optic cable with an error rate of less than $10^{-12}$. At the source wavelength ( 1.3 um from an indium-gallium-arsenide-phosphide laser at room temperature) the attenuation is such that the loss in the cable is only $0.73 \mathrm{~dB} / \mathrm{km}$ (including losses in the splicing, every 2 kms ).


## Bread Power

Continental Specialties Corporation has introduced a kit of parts for a solderless breadboard system with three regulated DC power supplies.

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We know from our mail bag that the UK edition of ETI is read all over the world. Recently we received our first fan letter from Russia (a red letter day?). The demand for ETI behind the Iron Curtain is so great that the USSR National Public Library for Science and Technology have asked us to send them a copy every month. I wonder if they have an ulterior motive. 'Here is the news for 1999 - this morning five thousand String Things swept across the Central Ger-
man Plains, dessimating Nato forces. British Centurion tanks were no match for Soviet air to ground flash triggers. Ground troops had been thoroughly trained on Ambush simulators. We believe their operations coordination computer is called Triton.' Just a joke, comrades.
In return for a monthly Moscow air-lift of ETIs, the library have kindly offered us our choice of over 500 Soviet periodicals. They are all listed in a handy handbook . . . in Russian. Now, we're not devastating decoders of Cyrillic script at the best of times, so, to put it bluntly (Anthony who?), Houston, we have a problem.

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45p $v$.

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SPST on $/ \mathrm{ff}$
SOp $\begin{array}{ll}\text { SPST on/off } & \text { 80p } \\ \text { SPDT } \\ \text { C/Over } & 85 \mathrm{p}\end{array}$ $\begin{array}{ll}\text { SPDT C/OVer } & \mathbf{8 5 p} \\ \text { OPDT } 6 \text { Teg } & 85 p\end{array}$ misiature Mon Loctding Push to Make 15p

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## Breadboard'79

To all of you who came along to the grand electronics extravaganza - thanks for visiting our stand. To those of you who didn't - why didn't you? Try to get along to see us next year.
ETI was hard to miss. Together with Computing Today and Hobby Electronics we occupied the biggest stand in the show, much to the dismay of one of our competitors who was displaying its wares from a cupboard opposite us.
We did a brisk trade in magazines. In fact we couldn't stock up the stand quickly enough to keep pace with demand. Apologies to anyone who wasn't able to get the back number of his choice on the day.
Did you see the Hobby Electronics' HEBOT going through its paces? Shame on you if you didn't. Notice how Remcon (from whom you can get everything except the electronic components) killed two birds with one stone by using HEBOT to carry information leaflets round the gathered throng. In fact, there was a profusion of throngs of all shapes and sizes throughout the week at all of the stands. Throng members could be seen spending their pennies and flashing their cheque cards for bags of components, HEBOTS, radiometers (I bought one myself), breadboards and books.


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## Standard Crystals

Thomson-CSF can now offer ovened crystal oscillators, type PMT P5, with ageing rates as low as $\pm 5 \times 10^{-10}$ at any frequency between 4 MHz and 12 MHz . You can choose the operating temperature and stability to suit the application.

The oscillators each weigh 180 g , measure $67 \times 60 \times 40 \mathrm{~mm}$ and are $P C B$ mounting Frequency adjustment is by external potentiometer or an
externally applied voltage. The oscillators feature a very low signal to single sideband phase noise ratio at the output, the typical noise floor being $150 \mathrm{~dB} / \mathrm{Hz}$. The initial power consumption, with a $12 \mathrm{~V} \pm 5 \tau$, negative earth supply is initially 8.4 watts, but falls to about two watts steady state at $25^{\circ} \mathrm{C}$.

The range of oscillators is available from Thomson-CSF Components and Materials Ltd, Ringway House, Bell Road, Daneshill, Basingstoke RG24 0QG.


## Pil Pan

For about $£ 90$ you could have an LCD digital multimeter with 30 ranges, including five capacitance ranges and a temperature range using an optional temperature probe namely the PAN 2000.

Other features include automatic polarity and overload display. A nine volt battery
brings 150 hours of digital viewing. With the half kilo package you can test 100 uV to 1000 V and 100 uA to 2 A DC , 100 uV to 750 V and 100 uA to 2 A $\mathrm{AC}, 2 \mathrm{k}$ to 20 M .1 p to 20 u and $-50^{\circ} \mathrm{C}$ to $125^{\circ} \mathrm{C} .30 \mathrm{kV} \mathrm{DC}$ is also available.
The PAN 2000 is availble from Precision Instrument Laboratories, Instrument House. 727 Old Kent Road. London SE15.



OOPS!

## The Beast <br> (Nov.)

Some readers have experienced difficulty in winding the required 77 primary turns on the coil of track-cleaner transformer TI when using 22 s.w.g. enamelled wire. In case of such difficulty, 24 s.w.g. wire can be used instead.

We are presently aware of no other corrections pertaining to this project.

## MCC Preamp (Jan.)

In January's issue we wrongly stated that the case for the Moving Coil cartridge project was available from Boss Industries. It is, in fact, supplied by West Hyde Developments. Pease use order code Classic II AJD.

## Deep Space Probes(Jan.)

Due to poor printing, page 87 of the Deep Space Probes feature last month is difficult to read. If you have an unreadable copy, tear out page 87 and send it to us. We will send back a fresh copy.


# ROBOT PIN-UPS 

The response to our recent rallying call to the robot-makers of the UK has been truly astounding. Photos and detailed descriptions of the electronic entities have been pouring through our letterbox ever since.

Necessity, being the mother of invention, seems to have been responsible for the creation of metal men to cope with all those tedious yet absolutely essential jobs - from cutting the grass to frightening cats.

Here are a few of the ingenious creations that have come to our attention so far. Keep sending them in, with as many snaps as your Brownie can manage. Even if your system isn't finished yet, let us know what you're up to.


ELECTRONICS TODAY INTERNATIONAL - FEBRUARY 1980


UR 700 is equipped with six channel radio and light seeking and line following functions. It can 'feel' its way around using a tactile sensor band around the base.


UR370 started life as a two channel Futaba radio and a pound of splintered balsa wood. Both the UR 700 and its brother, the UR 370 are designed by Joe Gillespie.
This car battery and lawn mower motor powered butier is built from just about everything bar the kitchen sink. It incorporates a lampshade and three huge stewpots. Cyclops 2's arms are driven by six inch hydraulic rams and his nether regions spin round and flash. Cyclops 2 was designed by Steven Brooks.

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# CASIO FX 502/FA1 

# Whatever you call this Casio marvel - computer or calculator - it's an attractive little beast! Beguiled by boundless buttons, bewitched by beautious boxes, belaboured Ron Harris reports. 

$T$ is considered good form to be able to define precisely what you are reviewing, before sitting down to churn out the hackneyed prose which will enrage the manufacturer and baffle the consumer. In this case, however, I must confess that I transgress. Why?

Consider this: a computing system with ten program registers, 100 user labels ( 10 per program), LCD alphanumeric display, cassette dumping, full conditional branching, random number generation, dynamic display control (one second 'PAUSE' delay-stackable) and an excellent number crunching facility. Sounds like whatever it describes is well on the way to being a good micro system does it not?

No, it does not. It sound exactly like the Casio FX502P/FM1, however. Else why would it be in this
review anyway?). Apart from all that the same machine is an excellent scientific calculator and is packed and presented so superbly that I can foresee large numbers of collar and tie impulse buyers trooping home to the missus, wondering how best to explain $£ 100$ spent on this calculator set-up that they just happened to pick up

The Casio cosmetics inspire one to heights for avarice hitherto reserved for expensive trifles such a cars, hi-fi and precious metals!

[^1]


## Whats In It For...

The 502 comes with two keyboard overlays - one for program use, one for music use of which more later) - a program library covering such diverse topics as Surveying and Transistor Amplifier Design ;but precious few games), an Operating Manual which is all one has come to expect and detest in such publications and a very smart wallet-type carrying case

The FA1 cassette adaptor also has a carrying case, manual and short-form operating guide. The construction quality of cassette and adaptor is very high indeed. All in all eleven out of ten for presentation, Casio

As a scientific the 502P has just about all the functions you are ever likely to consider using. Such statistical creations as standard deviation, mean square sum and data number are all hanging around beneath a key somewhere too, as are 22 non-volatile memories.

Mains operation of the 502P is not possible at all, but battery life is somewhere around 1300 hours ;of operating time) and an auto switch off will turn off the calculator if you leave it on, unused, for more than a quarter of an hour. Eat your heart out Electricity Board.

The display is nothing short of superb. Normally it shows ten digits with two digit exponent, but can be rounded to any number of digits, or switched to scientific, engeering or sexaouesimal (degree) mode. Neat little alpha-numeric prompts appear in the lowest $1 / 3$ of the display to keep you mindful of what you are doing with all this power.

Ten out of eleven as a scientific, Casio.

Right: a 502P fully dressed in its programming overlay. The little numbers beside each key are the check codes used by the machine (and you) in the compliation of a listing. The bottom half of the overlay is clear, to enable the multi-function keys to be read.

Left: the calculator nicely settled into its FA1 tape adaptor. Two slots in the holder engage the black lugs along the calculator sides to ensure a good alignment between the (gold plated) plug and socket.

## Big Steps For Man, 256 Steps For...

Before the 502P Casio used their own little FORTRAN based programming language. That is now gone to join the Dodo, and there won't be many mourners I think.

They have returned to a more straightforward assembler type approach, but have an operating system which puts them clear ahead of the opposition. The total 256 steps of program memory is divisible into ten routines, PO-P9, and each can be called as a subroutine within another program, up to four deep.

Puttig a program into memory could not be easier. Set the mode to write - display shows which programs are free - designate (choose PO-P9) and enter just as you would if carrying out the calculation via the keyboard SINCE there is no absolute addressing, all jumps must be labelled. Most of my programs began with 'LBL 1 ' just in case . . . . As you've got ten lables per program to play with - splash 'em all over! Correcting listings is easier than ever, the 502 can step forward and backward, single step or fast, and can delete or insert steps and automatically renumber the rest accordingly!


Conditional jumping is most comprehensivelyprovided. Six tests can be made $-x=0 ; X>F ; X<F$; ISZ; DSZ - and jumps made or not as required. The last two are ''increment and decrement jump if zero' ' commands which are particularly useful.

At each step the display gives last step-number and check code for key pushed at that point. Tis check code is alpha-numeric and corresponds to the programming overlay: After a few times through you get used to most of the codes and things (as always) improve with practise.

If the conservative 256 steps are not enough for you then the FA1 allows you to dump programs into a cassette player. Load times are pretty short so that over 100 programs could be lumped onto a C-60, PER SIDE!

## Measure Of Tape

The FA1 terminates in a set of twin jacks, which should fit most portable cassette players, but will require an adaptor to fit most hi-fi decks. A few minutes with the soldering-iron might be preferable, though.

Programs are given three digit file numbers for both SAVE and LOAD operations and the machine will search the tape until it finds the required number to load. The really nice touch about this is that programs and data are stored separately, so that different data can be run through the same program, and sub-routines extracted to be added to existing listings within the calculator. Frighteningly flexible.
This brings me on to the music playing facility of the 502 /FA 1 combination. I cannot ignore it any longer, I suppose. All I shall say is that I consider it an insult to the intelligence of anyone buying such a calculator. What does it sound like, you may ask. Horrendous I might say. Is it
entertaining, you might ask. Boring, I might say
Thus we could deal with the thing in one paragraph. Yeuk!

I found no problems loading and saving programs from the FA1 at all, and consider it a good system. Keep the volume turned up full on the tape through.

## Summing It All Up

Overall then the new Casio has a lot going for it. A beautifully elegant programming system and a good scientific ability. Coupled to a tape through the FA 1 it becomes an unequalleld way to get some real computing power into your life at a reasonable price. Also an excellent way to earn basic programming techniques, as the machine steers you toward modular programming almost instinctively.

Against it stands the monument to ineptitude that is the Instruction Manual. Please Casio, you sell yourselves short here. The 502P is a magnificent piece of work but why why why why make it rise against a manual like this?

Other minor quibbles - although the display has alpha-numeric capability it shows the check codes doesn't it) the user is denied access to anything but numbers. Why not let us loose upon the full ASCII set, eh? Make for much more interesting programs. And since we have a little brother (FX501P) with 128 steps, can I request a world conquering FX-503P with 1000 steps and absolute addressing please?

Our thanks to TEMPUS of Cambridge for supplying the review machines to us. Prices for the FX502P and FA1 are $£ 74.95$ and $£ 19.95$ respectively, although for a limited time TEMPUS will sell you the pair plus a Masterpack - a software pack of 150 programs on cassette - for $£ 99.90$ all inc. Tempus, Beaumont Centre, 164-167 East Road, Cambridge CB1 1 DB .

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The latest development to emerge from the semiconductor research establishments is the VMOS FET. This device seems set to rapidly revolutionise many areas of solid-state technology, particularly fast switching/high current applications such as line drivers, optical data transmission, memory drivers, DC/ DC converters and so on. However, VMOS devices also have unique advantages in communications circuitry. These are:

1. They can be used in any class of operation (A to $D$ ) due to the advantages of enhancement-mode operation.
2. They have excellent linear transfer characteristics. Fifth order inter-modulation distortion figures are typically eight to ten $d B$ better than bipolars
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5. They are, by virtue of their construction, protected against infinite VSWR problems.
6. They have a high source-drain breakdown voltage.
7. They have low gate-drain feedback capacitance. This characteris-

## SPECIFICATION

## Siturited output power:

Power Gain:

Efficiency:
Intiput VSWR:

Vd of 24 solts -76 watts ( 13.5 -wstts specified) Vd of 36 vols -22 watts ( 21 watte specifiod) Vd of 24 volts $-11,2 \mathrm{~dB}$ at 19 watts output (10 6 B @ 200 MHz specifiod) Vdo of 36 vol ${ }^{\text {d }}$ - 11 dd at 18 -watts oupu 53\%
lessitharr1.5:1.

The prota侮pe was unconditionally stable without neutralisetion.

tic particularly suits $\mathrm{VHF} / \mathrm{UHF}$ operation.
8. Low noise figure: 2.5 dB is typical. A power amplifier can be used as a low-level amp as well
9. Wide bandwidth and uniformity of characteristics across frequency range.
10. Higher gain than bipolars of equivalent power dissipation rating.

## HOW IT WORKS



11. Input and output impedances generally higher than equivalent bipolar devices making matching easier

High power MOS devices were pioneered in the mid-1960's by RCA laboratories who managed to produce an amplifier that delivered up to 14 watts at 10 MHz . The Russians next achieved 1 watt at frequencies up to 100 MHz .

Recently DMOS (double-diffused MOS) was developed in Japan and commercially produced by Signetics.

The performance of these early types of MOS technology has been surpassed by VMOS which can offer higher power levels by virtue of its inherent improved thermal transfer in the chip construction.

Several companies have developed VMOS devices for communications applications. These are: Westinghouse, Siliconix and the Communications Transistor Corporation. Other companies, such as Fairchild, have confined their interests to fast switching devices.

## Construction

Commence construction by drilling two holes diagonally opposite the transistor mounting hole, as shown in figure 2 and the photograph, to clear the mounting bolts for the transistor flange. A small file should be used to elongate the holes.

Drill two holes in the heatsink to accommodate the transistor flange securing bolts.

Next, carefully solder the VMP-4 to the copper side of the PCboard taking care that the orientation is correct:


Fig. 2. Component Overiay.

NOTE: The bias and neutralisation components are not shown on the overlay. Neutralisation was not found to be necessary on the prototype.

If bias is required for linear operation R1 should be fed to the bias supply and this point by-passed through C3.

## PARTS LIST



## Coil Winding Details

L1
5 turns 18 B + S (19 SWG) tinned copper wire 6 mm inside dia 10 mm long.

L2
3 turns $18 \mathrm{~B}+\mathrm{S}(19$ SWG) tinned copper wire 6 mm inside dia. 6 mm long

L3
4 turns 26 B + S 127 SWG) enamelled wire wound on Neoside L1010 former with F29 slug.

## BUYLINES

The onlv comporient liatle to the the intending constructor is the VMPT All the rest are Jairly stendard stulfe end people tike Cavonics, for ex anple, will have stocks.

The VFET itself is ovaifable from P. Rimmer at 367 Greêrrlane, N4 104 or Ambit Internationial, Check with them forg prices before ordering.
you have checked that all is correct, testing can commence.

A variable supply with a current limiting facility is suggested for initial test and tune up. Bias is not necessary at this stage

Connect the output to a dummy load and some RF power measuring device. Apply supply voltage and then drive power. Tune for maximum RF output! This should correspond with a peak in drain current. If 'funny' things happen here then suspect positive feedback and install neutralisation.

It's dead simple. However, take care not to grossly overdrive the device - VFETS do not take kindly to this sort of abuse. Becoming the owner of a four-legged stripline fuse can be a chastening experience!

For class $A B$ linear operation, bias should be set to provide about 100 mA quiescent drain current, subsequently adjusted for best performance

Some adjustment of the input network was required on the prototype suggesting that the input impedance, $\mathrm{S}_{11}$, of the particular VMP-4 was lower than indicated on the data sheet.


| MTFDiS. 1.5 V Madel moler 32 P , Sub. minin. Dit ach $115 V 0 C 3$ rom motor 32p. I2V DC 12 OC molors $55 p$, Geared mains molors. 2h rian. Thp. 115 NAC 4 rjm molor 95 p . IC Catente motors 95p EV Cossutio motors mive Andard lat. trpe $\mathbb{1} 1.25$ | RECORDTNG TAPE. Low mols mylat <br>  ETTES. C60 10 tor E2.75. CSO 10 tor $\mathbb{E} 3$ B TAPE ME DD DEWAGNETISER, 240vas curwd robs E2.65. LOW COST $1 / 2^{\prime \prime}$ tape splicer E1.95 | amprenol cannectons [pr 599$]$ <br>  52 ohms I wath with indicator bult 97 p . BHC plugs erimp type 38p each. | MULTIMETERS, NH55 2,090 EpY. IXV <br>  current resibnce lo 200 was in 4 rames. <br>  <br>  man, E34.50 | TR县NSFORDERS. All 240V 0C primary pasiage per iransiormar is stown in <br>  <br>  <br>  <br>  30 V 2 E 4.75 (54p) 20 Y 2.5 A E2 35 ( 34 p ). | alumimum pajuct raxes. 18 <br>  <br>  ${ }_{6 \times 6 \times 3 \text { 30. }}^{1.25}$ |
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# BUGGED BY INTERFERENCE? 

## You can banish the demon hum from your sounds if you know what you're looking for.

The most common irritant in hi-fi systems of any performance or cost standard is almost certainly hum, that horrible, low pitched noise which strikes almost every system at least once in its life. In fact, the only good thing that can be said about hum is that it is not selective; it will strike both amateur and professional equally.

And the most aggravating thing about hum is the trouble that it may cause - not that it will normally damage components unless extremely bad, but because it may sometimes take hours of concentrated searching to track it down. Until the cause is located and the problem corrected, satisfying listening is impossible.

## What is Hum?

Once hum has been heard it is unmistakable. It is also difficult to describe. Typical textbook descriptions run pretty much as follows: 'HUM - an unwanted low-pitched sound produced in reproduction by an interference from the AC mains. It usually occurs at the mains frequency of 50 Hz , or at its second harmonic, 100 Hz . Can be caused by ...', and then follows a list of about a dozen typical causes.

In spite of the rather open-ended descriptions, hum is immediately recognisable.

Its causes are not so easily pinpointed, although the sources of hum found in hi-fi systems can usually be traced back to any one of three chief sources - loops, screening and induction.

Hum generally affects low level signals with high circuit impedances - so the pickup cartridge and its associated signal connections are the prime offenders. The high gain of this circuit, and the large amount of bass boost applied in the RIAA equalisation, make it exceptionally prone to hum from any of the various causes.

In Britain, the AC mains power supply is a nominal 240 volts, alternating with a frequency of 50 Hz . Wherever these voltages occur, the mains conductors are surrounded by electrostatic and magnetic fields which fluctuate at the same frequency. With a voltage as high as 240 volts, these fields are fairly intense and can produce hum by inducing tiny $A C$ currents in surrounding wiring and components.

Occasionally hum may arise from a faulty component in some piece of equipment, but most problems come from the linking and positioning of the hi-fi components.


Hum-bugs can strike ANY system, with no regard to make or quality!

## Loops

Earth loops are possibly the most common cause of hum, and are the most annoying in that they are frequently caused by taking too much care! They are formed by duplicating earth links between components - a real trap for beginners attempting to set up a foolproof system.

The problem arises when the earthed screen of a signal carrying lead is correctly earthed, but an additional earth link is formed between the two components. The separate earth link may be redundant, in which case it forms an electrical loop.

If there are slight potential differences at the earth points, or if the loop falls within a stray magnetic field, a tiny circulating AC current will start to flow within the loop. This current affects the audio signal carried in the central core of the conductor by adding unwanted components to it, and hum results.


Sometimes the loop may be formed through no fault of the person assembling the system. When the earthy sides of the signal connectors on the amplifier (or other component) are linked internally to form a common ground, a loop is very easily completed.

Troublesome loops are known as 'earth' or 'hum' loops, and although they are the common cause of a great many problems, their source may well be different in each case, and their cure may take considerable time and careful thought.

The ideal interconnecting system between any two components conveys the signal and the earthy (signal return) conductors for each channel by only one path - via a live and an earthed conductor. Separate earth connections should be used only to earthed metalwork which is not connected to the signal carrying circuits. Because of this pickup arm on a turntable is earthed separately without causing hum loop problems.

In these cases earthing may be essential to draw off any leaked voltages or static build-up directly to the mains earthing point. If these spurious voltages were carried via the signal earth leads, they too could cause interference with the audio signal.

## Curing Earth Loops

As hum loops are set up when there is a redundant earth connection, they can be cured by breaking the loop - that is, by removing the redundant earth.

The process of tracking down a hum loop problem is rather long and laborious. If the majority of connections are made by RCA-phono plugs and sockets, it is rather more simple than when DIN connectors are used.

Start by disconnecting all inputs to the amplifier except for the separate earth link between the turntable frame and the amplifier's chassis. Connect one channel of the turntable and note the hum level. Try pulling the plug slightly out of the socket so that the outer rim doesn't still make contact. If the hum is now worse, push the plug fully home - excessive hum indicates that the rim contact is not redundant but is needed for shielding. If, however, the hum is reduced with partial connection, leave the plug as it is - this earth is redundant.

Now check the second channel from the turntable in the same way. If the hum is still present in spite of these checks, try removing the separate earth link from the frame of the turntable to the amp. It is unlikely that this will be redundant, as ideally it should be isolated from the signal carrying components. In rare cases, however, its removal may help.

Continue this checking process in the same way for all inputs, until you are sure which earth links are redundant and may be disconnected permanently.

DIN connectors pose more of a problem, and unless you are sure of your ability with a soldering iron it will be better to check only the auxiliary earth connections, and then investigate to see if the problem lies elsewhere.

When working through the checking procedure, make sure that the amplifier's volume control is turned down whenever making or breaking contacts. High level transients are easily generated, especially when checking for loops in the vicinity of the turntable.

> A the amplifier is the control centre of a system, central earthing is best carried out mt thin point to avoid loops. Only use single earth wherever possible

## Electrostatic Hum...

Because of the fairly intense electrostatic fields surrounding mains supply cables, any signal-carrying leads within such a field may be affected, because of the capacitance across the space between the cables. The higher the circuit impedance, and the lower the audio signal level, the more likely the occurrence of hum breakthrough.

This problem is generally overcome by the use of an earthed shield around the signal carrying conductors - hence the almost universal use of shielded leads for connections between components.

The screen must be arranged so that the live conductors are shielded by some earthed metal at all times - it will be seen when looking at RCA or DIN plugs and their appropriate sockets, that this requirement is fulfilled. It is to prevent this type of hum that almost all signal carrying leads between components in a hi-fi system use a shielded cable.

The shield, or braid, should itself be insulated so that it doesn't inadvertantly contact any other earthed metalwork. If this happens, the earth bypass created will form an earth loop to bring even more hum.

The only components which do not require shielded connecting leads for the audio signals are loudspeakers, which are fed by high level signals via a low impedance circuit.

## ...And The Cure

The most obvious cure for hum of this sort is prevention. Any signal-carrying lead should be kept well clear of all mains supply cables, and should also be kept as short as practicable. However, as electric fields are coupled by capacitance, and the effects diminish as distance increases, length should not be sacrificed unnecessarily. Never lengthen the connecting cables supplied with a turntable, however, as this will degrade the unit's performance.

Mains cables should consist of twisted conductors or, when using two core mains leads, parallelled conductors so that the fields are reduced by cancellation.

If it can be established that hum is caused by some form of electrostatic breakthrough, but it is not practicable to move the offending cables, a form of shielding may be required between the mains cables and the signal leads. Any earthed metal should service the purpose - provided any signal leads and thus set up an earth loop.

When turntable hum is the problem, a trick of the trade which works in a surprising number of cases (with turntables fitted with two pin mains plugs) is simply to reverse the two pin mains plug in the power outlet - whether at the mains or at the amplifier's mains outlet. By transposing the active and neutral conductors in this way, it is sometimes possible to reduce the field that may occur around a switch, or some


Turrtables and associated equipment are pertciulariy prone to hum in all its infuriating forms. This is due to the high impedances in this part of the circuitry and the bass boost applied by RIAA correction. Make sure for starters that
the turntable is not earthed twice through the signal leads and its own leads. Some cartridge manufacturers earth the metal of the body to one channel of the system and if you get mono hum in one speaker have a look at the back of your head! (cartridge wires therein)

While it may be an advantage to use earthed metal equipment cabinets as a shield against induced electrical hum, this is not possible when the cabinet is electrically connected to the equipment's chassis. In this situation, any separate earthing of the cabinet will form a hum loop - the cure may be worse than the original sympton.
other internal device which is sufficiently close to the pickup, or to signal leads, to cause problems.

It is also worth experimenting with different routes for the signal cables - keeping them well clear of any cables carrying mains voltages. Make sure that any mains conductors are kept well away from the pickup cartridge.


Hum fields exist around mains<br>transformers and AC<br>mains wiring in<br>general. If sensitive<br>equipment -<br>unscreened - is<br>placed neer enough<br>humbugs will<br>immediately surround<br>it making life<br>unbearable

## Magnetic Induction

Transformers and electric motors operate within powerful magnetic fields which are generated by passing the mains current through the windings of a coil. It is very difficult to contain that magnetic fields that occur around transformers or turntable motors, and they tend to spread out beyond the immediate vicinity of the device. Any coils (including earth loops) or windings used in signal carrying components which do fall within the stray magnetic field, are very prone to hum pickup of this type;

The earthed shield used to prevent hum from electrostatic fields is unfortunately no barrier to a magnetic field, and special metallic shielding - such as mu metal - must be used.

The components which are most susceptible to magnetic hum induction are the pickup cartridge and the magnetic heads on a tape recorder - low level devices which rely on magnetic coupling for their operation.

Generally tape heads are shielded by the internal design and layout of the recorder's electronics. However, the pickup cartridge by its very design and performance requirements must be close to the turntable's motor. While most good turntable motors do not give trouble, cartridges do vary in their sensitivity to magnetic fields, and troubles may occur when least expected.

In spite of any design features or in-built shielding included to prevent the breakthrough of magnetically induced hum in turntables and tape decks, these units should be kept as far as possible from the power transformers of amplifiers, tuners and other components.


## Curing Magnetic Hum

The most common problem with magnetically induced hum is found around the pickup cartridge. To cure this hum it is necessary first to establish the cause.

If you suspect that the turntable motor is the cause of the problem, try switching the motor on when the arm is at different points across the turntable. If the hum appears and disappears as the motor is turned on and off, then the motor is the culprit.

It will be possible to tell where the problem is worst, and the only cure in some cases will be to relocate the arm. If this is impossible, as it would generally be with automatic and semi-automatic arms, it is necessary to look to some form of magnetic shielding.

If hum levels change with the position of the arm across the platter - even with the motor switched off - it is probable that some nearby component is the cause - possibly a power transformer in the amplifier or some other component nearby. Try changing the position of the turntable relative to the other components - sometimes a slight change in orientation is all that is needed.

Similar experiments will be required if the hum occurs in a tape deck, although this is only likely to occur when the equipment is mounted in a confined space such as an equipment cabinet.

## Halting Hum

The search for the source of a hum may take a great deal of your time and when hum problems do arise, you must be prepared to devote several hours to the hunt. It is certainly worthwhile - you will not get any easy listening until it is found.

Remember that hum rarely comes from a faulty component so it is unlikely that you will banish it simply by buying replacement components.
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# HH-LO PASS FILTER 

# Are you in the depths of despair because your highs are hissy, or because your laws are rumbly? Then you need the ETI Hilo Filter to remove your extremities - without anaesthetic. 



We can hear you all now, saying to yourselves in unison simultaneously and all together - GOSH, is there no end to the ETI Project Team's ingenuity? Whatever will they come up with next? Well this really is the ultimate in audio gadgetry - a rumble and scratch filter that one can tune to suit an individual rumble or a personal scratch. Speaking for ourselves, we only scratch when there is no-one around anyway, it's bad manners to do otherwise

The ETI HILO FILTER essentially consists of two separate filtera, a variable high pass and a variable low pass filter in series, which together form a tunable bandpass filter, whose cutoff frequencies at each end of the audio spectrum can be adjusted by varying the cutoff frequencies of the individual filters.

The cutoff point of the high pass
filter is variable between about 235 Hz and 2.8 kHz depending on the setting of RV1. At all frequencies above this the filter operates as a unity gain buffer. The effect of this part of the whole circuit is to quite adequately filter out all low frequency oscillations and tremors which go under the heading "rumbles"

The low pass filter functions in just the opposite mode, in that it filters out all frequencies above its cutoff point ivariable between about 2.2 kHz and 24 kHz dependent on RV2).

Unwanted tape hiss and record scratches can be filtered out with this part of the circuit. Frequencies below the cutoff point are buffered with unity gain.

For the technically minded we can say that both filters are second order filters giving a cutoff rate of $12 \mathrm{~dB} /$ octave below or above the consecutive filter points.

## Construction

The circuit really is so simple that very little need be said on construction techniques. The only area of possible difficulty might be encountered in the wiring up of the two potentiometers RV1 and RV2. Both are dual ganged types and obviously require to be wired correctly for proper function. Work from the circuit diagram and follow the wiring carefully. If the circuit does not work first time, it is more than likely that this is the area of fault.

We build our filter into a box but this is not strictly necessary. You can build the board into your own system and run the filter from a mains / DC power supply ;up to $\pm 15 \mathrm{~V}$ ), which will save the expense of batteries, if you wish


The imemal photograph (left) shows the completed PCB installed in its case and the connectors to the batteries and front panel controls. Two PP3s will give sufficiont power for satisfactory operation. We have taken the outputs to two phono sockets on the back panel.

Fig. 1 (below) shows the circuit diagram. RV1 and RV2 are both ganged potentiometers.



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## HOW IT WORKS

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Sinitarly: the low pass thetre shows? Gequency respohtsp as in the graph of Fig. 2.-Adjasment of R 2 2 altery theycutoif frequncy 22 kitz and 24 kiz as shown by: the broken atid full lines of the werh.

When the two filery follow cach othatit as in ouf cirealt the overall xespanse curve is simply a combination of the two dlat rams of Thgs. 1 and 2.


You thouldn't find this PCB (above) very taxing to construct. It may be small, but the mighty mica morsel banishes hiss and rumble a treat.

Fig. 2 (below) shows the Hi-Lo Filter component overlay. The capacitors go in two by two - two polyester, two polycarbonate and two polystyrene.


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# CMOS 555 APPLCATIONS 

## Tim Orr brings you the result of bipolar versus CMOS in the 555 league

The bipolar 555 timer chip has been around for many years, but there is now a CMOS version that has some very significant design improvements. The two devices are functionally very similar, being interchangeable in most applications. The operation of the 555 is very simple, (Fig. 1). It consists of a pair of comparators that operate at $1 / 3$ and $2 / 3$ of the supply voltage, this being set up by a resistor chain. These comparators set and reset a flip flop which in turn drives the output stage. A second output is available which is an electronic switch, (Discharge) to ground. Otherl features include access to the resistor chain via the control voltage pin and an extra reset input to the flip flop. This simple network readily lends itself to all sorts of oscillators and timer circuits.


Fig. 1. The pin-out and internal configuration of the CMOS $\mathbf{5 5 5}$.
The bipolar 555 has a few parameters that can make it difficult to use, but which have been improved in the CMOS version, (Fig.2.). The bipolar quiescent supply current is generally about 10 mA which negates their use in small battery units. The CMOS version consumes a mere 120 uA.

| PARAMETER | ICM 7555 | BIPOLAR 555C |
| :---: | :---: | :---: |
| QUIESCENT CURRENT Vce +15 V | $\begin{aligned} & \text { TYPICAL } \\ & 120 \mathrm{uA} \end{aligned}$ | $\begin{aligned} & \text { TYPICAL } \\ & 10 \mathrm{~mA} \end{aligned}$ |
| INPUT CURRENT TRIGGER THRESHOLD RESET | $\begin{array}{r} 50 \mathrm{pA} \\ 50 \mathrm{pA} \\ 100 \mathrm{pA} \end{array}$ | $\begin{gathered} \text { Ou5A } \\ \text { 0u1A } \\ \text { Om1A } \end{gathered}$ |
| MAX. OPERATING FREQUENCY | 500kHz | 500 kHz |
| POWER SUPPLY RANGE | $2 \rightarrow 18 \mathrm{~V}$ | $4 \mathrm{VS} \rightarrow 16 \mathrm{~V}$ |
| PEAK SUPPLY CURRENT TRANSIENT | 10 mA | 370 mA |
| RISE AND FALL TIME at output | 40 nS | 100hS |

Fig. 2. A comparison between the Bipolar and CMOS version of the 555.

Also the CMOS inputs are very high impedance having input currents down in the pico amp region. Another major improvement is the reduction in the power supply current transient during an output transmission. The bipolar is very noisy in this respect and can often be the cause of lots of 'funnies' in nearby circuits.

The CMOS 555 is a low power high input impedance device that should be used where low current consumption is at a premium. The following circuits illustrate some possible uses of the device.


Fig. 3. The CMOS 555 displays an impressive reduction in supply current transient during an output transmission.


Fig. 4 (above) A simple oscillator can be constructed using two resistors and a capacitor. Due to the high input impedance of the CMOS device, resistor values up to 100 M may be used. The operation is as follows. Capacitor $C$ is charged up via $R a$ and $R b$ and $s 0$ rises with a time constant of $C(R a+R b)$. When the voitage $B$ reaches $2 / 3$ Vcc, the threshoid hold comparator goes high causing the output (pin 3) to be set low. Also the discharge FET (pin 7) is turned on. This discharges the capacitor via Rb with a time constant of CRb, until the voltage $B$ reaches $1 / 3$ Vcc.

This causes the trigger comparator to go low which sets the output high and turns off the discharge FET. Thus, the voltage B oscillates between $1 / 3$ and $2 / 3$ Vcc. The waveform is asymmetric but due to the nature of its generation its frequency is virtually invarient with regerds to supply voltage changes. It is possible to generate a sawtooth waveform by reducing Rb to a short circuit. This causes the reeet time to be very fast, of the order of a few microseconds. However, there is a propagation delay through the device from the trigger comparator to the output which causes the discharging waveform to overshoot the $1 / 3$ Vcc limit, thus making the waveform at B larger. This will cause the oscillation frequency to be lower than calculated. To maintain the calculated frequency, the discharge period should be 5 uS seconds or longer.



Fig. 5 (above) This oscillator has a symmetrical output because the charge and discharge paths are the same and the portion of the exponertial cruve that is used is symmetrical. Note that in this circuit, the discharge pin (7) is available to do other jobs, such as drive a LED or some other device. The timing resistor $R$ should be kept relatively high (above 10k) to prevent loading of the output.


Fig. 6 (above) This circuit is an oscillator that has a wide mark to space ratio. It is OFF for approximately 1 second and then turns ON for about 10 mS . During this latter period the LED is turned on by the discharge FET. The long OFF period is generated by the R1, C1
time constant, whereas the short ON period is produced by D1 (which is forward biased) and R2, C1. The average current consumption is relatively low. If the unit were powered from $\mathbf{9} \mathbf{V}$ battery, then the current would be 120 uA for the ICM 7555 and an average of 140 uA for the LED, making 260 uA total. This would give a lifetime of a few months for a PP3 which would be extended by lengthening the OFF period (increase R1) and reducing the LED current (increase R3).


Fig. 7 (above) The 7555 is used to initiate and terminiate a triggered sweep. Nomally the discharge FET (pin 7) is ON and so C1 is shorted to ground. When a trigger is applied to the input, the collector of $\mathbf{Q} 2$ momentarily goes low which sets pin 7 of the $\mathbf{7 5 6 5}$ in its OFF condition. Q1, R1, R2 and D1 form a current generator that drives C1. Once the discharge FET has been turned OFF, the voltage on C1 rises linearly. When this voltage reaches $+2 / 3$ Vcc, the threshold comparator sets the discharge FET into its ON state and so C1 is shorted to ground. IC1 is used to buffer the voltage on C1. The sweep generator is not retriggerable and is only initiated on fast positive going inputs. To vary the sweep rate, after C1 and or R2.

-śv no load
-TVAT 0.5mA

DC TO DC CONVERTER


Fig. 8 (above) A DC to DC converter can be constructed from an oscillator and a diode charge pump. The 7555 forms a high frequency square wave oscillator. The squerewave from pin 3 is AC coupled via C2 to the charge pump. The voltage on the negative side of C2 is preverrted from going more than OV7 positive by D1 and so the square wave on this side of the capacitor biases itself so that it moves from +0 V 7 to -8 V 3 . D2 charges up C3 on the negative excursion of this waveform and so a negative rail of about -8 V is generated. The current that can be taken from this rail is rahter low, being determined by the oscillation frequency and C2. Generally DC to DC converters have a poor transfer efficiency.



Fig. 9 (above) The 7555 can be used as the driving oscillator in an ultrasonic remote control system. The oscillator generates a thin pulee about $\mathbf{2 . 5}$ uS long at the natural resonant frequency of the transducer. This short pulse is used to turn on a transistor (a1) which drives an auto transformer with a 10 to 1 step up ratio. The output of the transformer is connected to the transducer and when the oecillation frequency is correct a 100 V peak to peak sine wave will be produced at this point. The transducer is usually a crystal with a high impedance and so a high operating voltage is required to produce any power output. The receiver is 040 kHz bandpass filter. This will amplify any audio signals at this frequency, which can then be sent to a detector circuit.


Fig. 10 (above) This oscillator allows the mark space ratio at pin 3 to be varied from 1 to 20, to 20 to 1 , by using two feedback routes. When pin 3 is high, C1 is charged via D1, part of RV1 and R1. When it is low, C1 is charged via D2, the other part of RV1 and R1. Thus thie position of the RV1 wiper determines the ratio of the charge and discharge periods. The oscillation frequency is slightly dependent on this ratio.

Fig. 11 (above) By making a loop out of a 7555, an inverter and an integrator, a triangle/square wave oscillator is produced. To operate well at high frequencies (up to $\mathbf{4 0} \mathbf{k H z}$ ) a CMOS inverter should be used to replace the transistor inverter. The op amp provides a low impedance triangle output, the frequency of which may be controlled with the 1 M log pot.


Fig. 12 (above) The 7555 can be used to generate an acoustic tone. The oscillator is set to run at $1 \mathbf{k 3 H z}$ which has a low period of about 15 nS and a high period of about 755 nS . During the low period the transistor is turned ON and so the loudspeaker is connected acrosa the power supply and sinks at about 100 mA (for a 9 V supply). This gives it a 'kick' on every cycle of the oscillation. As the transistor is only on for 15 out of every 770 nS the average current through the speaker is quite small, being about 1.95 mA . Therefore, the total current consumed by the whole system is only about 2.5 mA (at 9 V), and yet the $\mathbf{2 0} \mathbf{~ m W}$ output signal is quite audible.


Fig. 13. A warbling tone can be generated by using two oscillators. The warble is produced by IC 1 which generates a 13 Hz wave form thet is used to frequency modulate the tone generator as described in the previous example. Pin 5 of a 7555 is connected to the $+2 / 3$ Vectup on the rasistor ladder. By tying it to a 'warble' waveform, frequency modulation of the final output tone is produced. A 7556 could be used instead of two 7555's.


Fig. 14 (mbove) The police and other emergency services use a repeating high frequency 'beep' on their radio networks. This doesn't interfere with the normal radio traffic and allows the listener to be certain that he is still tuned in correctly to that channel. The circuit generates a similar 'beep' and yet consumes only a couple of milliamps. IC 1 is a slow oscillator ( 3 second period) with a large mark/ space ratio. The discharge FET is on for most of the time and only goes OFF for about 15 milliseconds in every 3 seconds. This FET is connected to the tone generator in such a manner that when the FET is ON the tone generator is inhibited. When the FET goes OFF the generator produces a burst of $\mathbf{3 k H z}$ oscillations which are heard as a 'beap'.


Fig. 15 (above) The last type of sound generator to be described is a simple siren. IC1 generetes a sewtooth waveform which is used to frequency modulate, via pin 5, the tone generator (IC2). As the sawtooth voltage rises (with a period of 1 S) so does the tone generator frequency.


Fig. 16 (above) The 7555 can be made into a monostable, although some problems may occur in its use. A negative going pulse on the trigger inpurt (pin 2) can be used to start the monostable period. It is important that this pulse goes high again before the end of the monostable period, or else it may prolong the period. To this end an AC coupled transistor inverter has been used so that a rising positive signal will initiate the event. Initially C2 is discharged to ground. When pin 2 is taken low, the discharge FET is sot to be OFF and so the voltage at E rises with a time constant of C2, R3. When this voltage reaches $+2 / 3$ Vcc, the discharge FET is turned ON, C2 is discharged to ground and the monostable period is finished. During this period the $\mathbf{7 5 5 5}$ produces a high output at pin 3.


Fig. 17 (above) The provious circuit has been modified. The timing resistor has been replaced with a programmed current source, IC2. Whatever current is put into pin 5 of the CA 3080 (the I will appear at its output. This will linearly charge up C2 when the FET switch is turned OFF. The monostable action will be the same as in the previous example. The monostable period is linearly proportional to the $I_{\text {ABC }}$ current so by programming this current the period is controlled.



Fig. 18 (above) The monostable can be used to linearly convert a frequency into a voltage. The pulses at the output pin of the 7555 in this circuit are fixed in length but occur at the fundamental frequency of the input signal. Thus the average equivalent DC voltage of these pulses will be linearly proportional to their frequency. This averaging can be performed very simply with a CR low pass filter. This filter determines the response time of the circuit and the ripple. A long time constant will have little ripple and respond slowly and vice verse. Care must be taken to not exceed the range of the monostable. When the period of monostable exceeds thet of the input signal, the circuit will miss every other period and will drop its apparent output by an octave.

Fig. 19 (right) IC1 and 2 form a triangle/ squarewave oscillator. IC2 is a non-inverting, progremmable rate integrator and its output ramps up and down between the $1 / 3$ and $2 / 3$ Vcc limits set by IC1. The rate at which the integrator ramps up and down and hence the oecillation frequency, is determined by the $I_{A B C}$ current. Because the input impedance of IC1 is so high, it is possible to directly connect the timing capacitor to it, without any unwanted loading effects. This circuit has excellent high frequency performance producing good quality triangla wave forms at 40 kHz . The circuit can also produce very low frequency signals by making C a 100 n tantalum and $I_{\text {Asc }}$ a current of say, 1 nA. This oscillator would then have a


 8
period of 800 seconds. IC3 is used to buffer the triangle to the outside world. If a 741 is used, then the waveform will become asymmetrical at low $I_{\text {ABC }}$ currents (below 1 uA ) due to the input bias current needed to run a 741. A TL081 has very little input current and so could be used for both low $I_{A B C}$ currents and high frequency performance. If low power is needed, then a TL061 (200 uA quiescent) could be used.

By closing switch S1, the oscillator will produce a sawtooth waveform and a pulse.




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Only a couple of cheap chips are used along with three transistors and a scattering of discrete passive components. Depressing the push-button causes the unit to emit a whooping siren sound. Also the LEDs flash alternately, gradually slowing down until one remains illuminated. A further depression of the push-button starts the process again. If the unit is left undisturbed it will turn itself off, emitting a couple of quiet squeaks. Once the flashing display has settled and one LED is illuminated continuously, the display will remain stable until both LEDs are extinguished. This avoids any troublesome arguments!

## Construction

The unit is assembled on a single PCB with the battery, push-button switch, LEDs and loudspeaker connected via flying leads. There are two links to make on the board; from IC1 pin 4) to IC2 (pin 4) and from IC1 (pin 3) to the junction of diodes D6, 7. Note that this is a straight wire link which

The finished coin machine (right) installed in a standard Verobox.



Fig. 1 Circuit diagram. We found that the circuit operated more reliably if D1 was made a 1 N4001.

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The Coin Tos project uses stemdard etmapotents. You should heve no affeulty in obtaining them from your fvourita mail order supplier:
passes under R11 and D6. The flying lead link is most easily made on the underside of the board after construction is complete.

Assembly is quite simple. As usual, pay attention to the orientation of the diodes, transistors, ICs and capacitors. C9 may have to be bent to one side to accommodate C1. However. there is plenty of space for the other components. As IC1 is a CMOS chip, use a socket if you are worried about damage from static electricity. There are no problems with IC2 or the other components.

Layout of the unit is uncritical and any construction technique may be employed. You need two different coloured LEDs. Any size or colour of LED will do. The unit should work as soon as power is applied and again upon depressing PB1. Build one now and try your luck!

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Fig. 2. Component overlay.

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# This month Dave Raven of Metac Electronics goes computer shopping and casts doubts on lie detectors. 

Starting a new year is not only necessary but can be quite enjoyable since it now possible to reflect on the previous year as history. Errors of judgment, success and missed opportunities are all behind us and cannot be brought back. The only positive thing you can do is cast all thoughts and aspirations forward. In the words of G. B. Shaw "Man can climb to the highest summits but he "cannot dwell there long"

If electronics is your chosen career then 1980 is to be another exciting year of challenge and new opportunities. Behind clouds of doom and gloom painted for us by the Press and politicians, not forgetting those great British patriots, the government and business spokesmen, who roam the world speaking out, never failing to tell the world how bad things are in Britain or how bad they are going to be, there are still signs of the continuing revolution in the electronics industry. Difficult decisions have to be taken, since, to remain in the industry, it will be necessary to spot the trends. An interest in small computing systems must be formed since organisations that wish to remain competitive and efficient have to make the change from mechanical to computerised systems. Fewer jobs, even for school-leavers, will be available to those who have not had sight of a computer keyboard. These new tools for industry and commerce are being pushed out onto the market at an alarming pace. It is already possible to buy a small system with 4 K of memory for half the price of a modern shop cash register. My own company is introducing a mixture of small systems for general purpose operations backed by larger units for the mail order centre. This philosophy of decentralising computers away from one large system gives maximum flexibility and enables a company to match the computer system to the job.

In America it is not uncommon for solicitors to sit at their word processor and prepare complex company merger documents in a fraction of the time previously required. So watch out ladies, this is just as important for you ton. Secretaries and office juniors will all need the ability to change from typing pool mentalities to the new 'smart' methods.

## Computers In The High Street

Major efforts have been made in the UK during 1979 to bring small computers onto the High Street. However, my own impression is that most of the retailers have ended up selling mainly to established businesses. The consumer will
not be rushed into these things and it does appear that home computers have still to take off. I would also hazard a guess that video recorder specialists are finding a similar pattern of sales. Currys, the electrical retail chain, is to establish a subsidiary company to market small computers and provide computer programs. Operations are expected to start next year with outlets being opened at Currys established premises. They are aiming for the small to medium size business market, also industrial and commercial outlets. Figures quoted in the Press suggested that they aim to take a whole $10 \%$ of the domestic market in their first year which doesn't leave a lot for the rest of us. Coincidently, news that the president of the electronics giant Fairchild Camera and Instrument Corporation had resigned reminded me of an announcement he made several years ago predicting that their range of Time Band LED watches were going to take some huge percentage of the world's market, I think it was about $40 \%$. Not all that long after this they announced getting out of watches completely and promptly moved their entire UK service facility to somewhere in West Germany.

## Confusion In The Shops

An interesting article on Computing Today describes the experiences of the author when shopping in London's Tottenham Court Road for a small computer. I mention this since it provides a classic story of the current state of immaturity and confusion that exists in the High Street. His experience when buying a computer may not be dissimilar to those buying electronic watches, video recorders, etc, mainly a bewildering choice of different systems, functions and, of course, prices. It is quite possible to spend $£ 3,000$ on a computer and find that the system is totally unsuitable for your requirements. Shortage of software for particular systems is a common complaint and the price of the add-ons.

## Honest Truth Machines

A sinister new electronic product is being sold in America, called the Truth Machine. I have not seen these advertised in the UK yet, but when they come I can imagine a similar reaction to the one created in the States. Doubts have been cast on their accuracy. However, this has not prevented these solid state, voice stress analysers being used in personnel offices and other areas where people may stray from the truth. One leading professor of psychiatry and psychology at the University of Minnesota claimed that
lie detectors are about as accurate as flipping a coin. Professor Lykken goes on to say that you cannot discriminate with a stress analyser between a stress and a non-stress situation. Even if tests could detect stress, one could not assume that stress indicates lying. Speaking the truth under stressful circumstances would typically show up as stress in these machines. A more traditional machine used for lie detection, the polygraph, was compared for accuracy with one of the new stress analysers. Results showed that the stress analyser detected lies $38 \%$ and , 20\% of the time compared to the polygraph machines record of $92 \%$ and $76 \%$ assuming of course that stress indicates lying. The companies marketing the products will not disclose the identity of the integrated circuits used, saying that patent rights are still being sorted out. To use one particular model of stress analyser the operator first adjusts the circuitry to recognise the fundamental frequency range of the voice of the person under test. While talking to that person under non-stress conditions, .the user tunes the analyser with a one-knob control, until the number 20 registers on the LED display. That number


The Truth Machine - currently being sold in America for about $£ 75$ each. Doubts are cast on the accuracy of its results.

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signifies that the machine is locked into the fundamental voice frequency range. The claim then made by the manufacturers is that when a person experiences stress, normal harmonics disappear from the voice and they shift so that they are no longer integral multiples of the fundamental. A sensor which also doubles as a filter and may be a tuned circuit, is programmed to anticipate harmonic changes, and measure their duration and intensity. The LED display indicates stress by a rise in the numerical value at displays. The model described has a built in microphone, or it can be connected to an external mike or to a telephone. On the drawing board is another model which I hope my company will be cautious about handling. It incorporates a stress analyser in a digital watch. Just think of the reaction when you're told "The time is 9.30 and you're already lying through your teeth.

## New SAW Devices

One old saying which regularly comes to mind is that 'there ain't nothing new' especially in science. Acoustic surface wave devices are becoming more familiar and I was interested to see an article recently describing a new high speed SAW device on gallium arsenide material. The material used is the same as that required for producing Field Effect Transistors and, by making use of the properties of gallium arsenide, strain-sensitive FET heterodyne mixers can be fabricated in the epitaxial layer grown on the surface of the bulk gallium arsenide compound, producing a useful programmable tapped delay line

Well, my news for the researchers is that a research programme I worked on ten years ago at Birmingham University developed working devices as described above Ah well, back to the drawing board.

ET


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[^2]
# Electronics 



Unless you've heard it you will find it almost impossible to believe, American radio that is. Rick Maybury hot foot from Los Angeles brings back a report of one particular AM / FM station, KEIY. In a city that has over 80 other stations catering for just about every taste KEZY manages to capture a very large audience. Find out about this and the American radio scene in general in next month's issue.

## INTO <br> CONSTRUCTIONAL ELECTRONICS

Are you thinking about taking that first step into the wonderful world of electronics? Ian Sinclair begins a major new series just for you

Many newcomers are often daunted by the thought of building that first project. This series is designed to take you gently by the hand, through the jungle of jargon and coloured bits and pieces and hopefully leave you with the confidence and expertise to tackle just about anything. Much of the series is centred around the Eurobreadboard so why not drop a few hints to Father Christmas so you can be ready


## KIT REVIEW SPECIAL

Two kits this month. Even though we've delayed the Radio Control system for a few months we will still be presenting the review of the Servo Kit we hope to be using with the system.

Our second kit comes from THE kitmakers Heathkit. They may be a bit pricey but they are the best. The one we're reviewing is probably one of the finest pieces of test equipment we've ever seen. It's the Digital Multimeter with an LCD readout. If you think a price tag of around $£ 80$ is a bit steep then see what we think next month.


INFRA RED CONTROL
Look out for a new remote control system using Infra Red techniques. Coming next month.

## The February issue will be on sale January 11th

[^3]
# MODULAR SYNTHESISER 

## The synthesiser you've been waiting for. Presented as either a full spec., rack mounting, stage instrument or as a set of stand-alone modules, ETI's Project 80 gives you the options. Circuit designs by R. C. Blakey.

0ne of the fascinating, and sometimes irritating, aspects of electronics is its rate of technological growth. It is the recent development of customised integrated circuits for music applications which justifies another synthesiser project. The use of these devices greatly simplifies construction of stable, accurate and reliable modules, which will be of particular benefit to constructors. These new integrated circuits are equally useful to the musician since they allow greater control of many of the complex 'musicmaking' parameters.

## Design Philosophy

The popular conception of a synthesiser is a musical instrument which can produce all manner of wondrous sounds. In fact a synthesiser is a multi-function machine having sound generators, modifiers and mixers which are electrically compatible with one another. It is the designer who connects the various modules together in a particular manner to produce an integrated musical instrument. He (she) has certain targets of portability, ease of playing and cost -
and inevitably these result in compromises in terms of the capabilities of the instrument.

The basis of the current project is that for hobby application and greater exploration of musical expression there is no substitute for a truly modular system inter-connected by patchcords. Of course, patchboards or switches are alternatives to lots of drooping cords, but sooner or later the compromises begin as the cost, or the trouble, of adding more switches or boards arises. Furthermore a patchcord synthesiser is just as easy to play as a pre-set machine.


## SPECIFICATION

## empral Specification

The design features of the modules follow the principles adopted by many of the major manufacturers of synthesisers. These are:-

1. Exponential völtage control. For the basic signal and treatment units (oscillators and filters) this will be the widely used one volt per octave relation. ship. For some modules the response Tivel may be "variable, as with the voltage controllediamplifier.
2. 10 V minimum control range proMiding a spant of 10 octaves without switching.
3. 10 V peak to peak signals. Signal and control inputs via a hing op-amp stage. This tlow
multiple inputs. It is also aimple matter to alter the inputs to make the modules compatible with existing equipment.
4. Normally all inputs will have an input impedance of 100 k or a minimum of 47 k . This allows a single control sig. nal to drive several modules without sdverse loadingeffects.
5. Output impedance will be $1 k$ wherever practical, that is, without add. ing significantly to cost. These outputs may be mixed by shorting them toge. ther which results in an average of the. mixed signals.
6. DC coupling, except for a few instances whêre it ixis essential tơ avoid errors which may arise from any $D C$ drift and when the use of AC coupling
does not result in foss of flexibility.
7. $\pm 15, V$ supplies, common to all modules. Single PSU required.
8. All mor features to have voltage control capabitity for example, 0 con trol of filters, pulse width modulation. envelope generation and so on.

An important feature of the above is that it allows anything to be plugged anywhere without fear of causing damage. Where modules are interfaced with oxternal equipment then Yinch jack cockets are used to avoid problems which may arise from incompatibility. Another aspect that should be noted is that because one signal can control severnt modules any attenuation of the signal should be made at the input to the modules.


Above and below: the VCO and PSU modules as they appear in their free standing cases. Note the vernier drive on the PSU. This is for calibration of oscillators etc. This case will be standard throughout the series.


## Modules To Come

The modules to be described in this series are as follows:-

1. Power Supply: $\pm 15 \mathrm{~V}$ regulated and trimmed supplies which, as described, are sufficient to drive all of the proposed modules.
2. Voltage Controlled Oscillator. Good exponential response over a range from 10 Hz to over 16 kHz . Triangle, sawtooth, pulse and sine outputs of 0 to 10 $V$ amplitude and triangle and sine outputs of $\pm 5 \mathrm{~V}$ amplitude. Manual and voltage control of pulse width duty cycle from 0 to 100\%. Linear frequency modulation input. Three techniques for synchronising oscillators.
3. Voltage Controlled Low Frequency Oscillator. Same features as the VCO but with a frequency range of 0.2 Hz to 205 Hz with a 10 V control input. The frequency range extends to over 5 kHz with higher control voltage or by manual adjustment of initial frequency. The lower frequency limit is easily altered to suit individual requirements.
4. Processor. The flexibility of the input-output structure described above has one main disadvantage, namely, that to take full advantage of it one should have several output jacks for each control source and an attenuating potentiometer on each control and signal input. This can be overcome by having a pane!
(or panels) which only contain attenuating potentiometers and/or sets of commoned jack sockets for distribution purposes. Such panels are not powered and do not require further description. On the 'processor' module facilities have been incorporated for inverting two of the inputs and also for slowing down control signals by means of a lag circuit. The latter circuit may also be used as a control source by using a foot pedal.
5. Voltage Controlled Mixer. Four input mixer with manual and external voltage control of mix together with a single control for manual and external voltage control of pan.
6. Voltage Controlled Filters. A state variable filter with low pass, high pass, band pass and notch outputs will be described. Also a 24 dB /octave low pass filter which is the most widely used type for music applications. Voltage control of Q provided.

## 7. Voltage Controlled Envelope Shaper.

 Voltage control over the attack, initial decay and final decay responses of the envelope generator provides opportunity for more realistic envelopes, for example, a change in envelope with pitch. A third generation IC will be used which has many other features allowing control over the formant of the sound.8. Dual Voltage Controlled Amplifier. Exponential and linear control for increased flexibility and variable gain of both. Ability to partially bury the envelope so as to reduce the exponential final decay of the envelope which in imitative synthesis can be a problem. Provision for stopping the VCA so that when used in conjunction with other instruments it does not continue playing when the rest of the group has finished.
9. Dual Ring Modulator. Based on a single IC and having a wide dynamic range and high signal to noise ratio.
10. Noise Generator and Sample \& Hold Module. White, pink and red (random) noise sources. The sample and hold unit allows creation of sounds from a variety of sources.
11. External Input. Pre-amplifier to increase external signals to a level compatible with the other synthesiser modules. Includes an envelope follower with a variable threshold level to generate trigger pulses.
12. Power Amplifier. Two speaker channels and headphone output. It is not advisable to use the domestic $\mathrm{Hi}-\mathrm{Fi}$ for synthesisers since the tweeters can be destroyed by continuous tones at an output above a few watts. A pair of good quality speakers is a worthwhile investment. A separate power supply is used for the power amplifier.
13. Keyboard Controller. Digitally scanned keyboard capable of controlling a five octave keyboard. Sample and hold with portamento. Precise octave shift over range of -2 to +3 octaves. Gate and trigger pulses. Pitch bend with variable bend level. Suitable for standard monophonic synthesiser and microprocessor compatible for polyphonic capability.

## Construction

Each module is designed for both panel mounting or for housing within a standard low cost plastic case. The use of a case serves a number of purposes, one of these being a neat packaging of the modules for those who wish to construct them for other applications. The cased units are also appropriate for teaching and learning purposes, for example, the beginner should start by thoroughly exploring the functions and capabilities of each module. If electronic music is approached in this systematic way it becomes a relatively simple matter to create specific sounds at a later date. Furthermore one should not have any pre-conceived notions on the format of a synthesiser and a methodical study will enable the user to determine his approach to electronic music.

The panels are $228 \times 76 \mathrm{mms}$. $19 \times$ 3 inches). This size is widely used by commercial manufacturers of patchcord synthesisers. Such a panel will comfortably accommodate at least six control potentiometers or rotary switches with control knobs of an easily manipulated size. Standard quarter inch jack sockets are used on most of the professional equipment but their physical size limits the number of inputs and outputs per panel and it is back to compromises again. This project uses 3.5 mm jack sockets and up to twenty may be fitted onto the panel along with the six rotary controls. These miniature jack sockets should be of good quality. Screened patchcords are used to reduce the likelihood of crosstalk or noise. A major

advantage of panels is their ease of construction which also applies to the case to house the panels. 1.2 mm aluminium sheet provides adequate strength and a professional appearance can be obtained by spraying and the use of transfers for the markings. The cabinet can be constructed from a proprietary laminated blockboard assembled with 'Lok-Joint' fasteners. With this type of shelving a maximum of twelve modutes per row is advised unless intermediate supports are used. 9.5 mm hardwood beading along the edges of the shelves allows the panels to be mounted using small woodscrews and if only a small pilot hole is made then the panels can be repeatedly removed without refilling the holes. The panel approach provides flexibility in lay-out and virtually unlimited scope for expansion.


MODULE 1: POWER SUPPLY

A
$\pm 15 \mathrm{~V}$ power supply is used for the modules except for the power amplifier which has its own supply. A design based on the 723 regulator is used and the circuit is shown in Figure 1. The circuit incorporates the recommended features for stability and both rails can be adjusted to a precise 15 V output.
Construction
The PCB has been designed to suit the mounting holes in the recommended case and these same holes can be used
for vertical mounting of the power supply in the keyboard housing by means of L brackets. A $100 \mathrm{~mm} \times 65$ mm heatsink provides generous heat dissipation from the plastic box and this is secured to the outside of the rear panel. The transformer should be bolted onto a small piece of aluminium plate which is screwed down to mounting lugs moulded into the box. Addition of plastic feet to the box will ensure adequate ventilation through the grills.

Before connecting to the mains supply ensure that all mains connections

Fig 1 (below). Circuit diagram for the Project $\mathbf{8 0}$ power supply.


Above: both versions of the VCO, panel mounted for rack fixing. Note that all inputs/outputs are at the bottom of the panel to prevent wiring fouling the controls.

HOW IT WORKS - PSU
ysory

Power Supply

- A synthesiser requires very stable power sup. plies because in most designs the voltage is used to establish reference currents in the T2 exponential generators. It is also desirable that the voltages are set at the precise 15 V levels. The $\pm 15 \mathrm{~V}$ power supply for this project is based on the 723 voltage regulator. The circuit comprises two identical positive 15 V supplies with one of the outputs tied to the ground rail of the other to generate the $\pm 1 S V$ required.

The circuit will supply up to 300 mA per rail at full voltage and the current sensing resistor R 2 (R7) limits output to about 450 mA under overload or short circuit conditions, R3, TP 1 and R4 allow precise adjust-
ment of output voltages; R1 improves temperature stability; C3 increases ripple rejection; C4 is for compensation; and C5 reduces noise on the output which originates from the voltage reference diode in the IC.

The output is adequate for the basic project but if extension is envisaged then the same circuit board can be used for much greater putputs The only changes required are ${ }_{\text {in }}$ increased power rating of the transformer; appiopriate increase in value of fuses; adjusting R2 (R7), i.e. halving this resistor will double the current limit and the PCB has provision for two resistors in parallel to facilitate obtaining the correct value Wirewound 2WSReristors are also used since low
ohmic values are more easily obtained and the resistors are physically smaller than $1 . W$ carbon types; increasing power rating of transistors and when more than 1 A per rail is required the latter should be mounted on separate heatsinks.

Setting up voltage controlled synthesiser modules requires a voltage source that can be quickly adjusted to a precise value. A ten turn potentiometer which can be switched to positive or negative voltage is an asset This is shown on the boxed module. Two commoned outputs are provided, one for the output to the unit being calibrated while the other may be used for direct connection to a voltmeter. Installation of this variable voltage source is, optional and components are not listed.
to plug, switch and transformer are properly insulated. The calibration consists of setting TP 1 and TP 2 to obtain, as accurately as possible, +15 V and -15 $\checkmark$ respectively.

## PARTS LIST - PSU

## RESISTORS All $1 / 4 \mathrm{~W}, 5 \%$

R1,6 $\quad 1 \mathrm{ks}$
R2, 7 1R5, 2.5W wirewound
R3,8 $3 \mathrm{k3}$, metal filh, 100 ppm T.C.
R4,9: 3 KO , metal film, 100 ppm T. $\mathrm{C}_{4}$
R5, $10 \quad 1 \mathrm{k} 2$
TP 1,2 1 kO cermet trimmer
CAPACITORS
C1,6 $\quad 2200 \mathrm{uF}, 35 \mathrm{~V}$ electrolytic
C2,7 $\quad=100 \mathrm{nF}$, polyester
$\mathrm{C} 3,8,2 \mathrm{u} 2,25 \mathrm{~V}$ tăntalum
C4,9 - 1 nF, polyester
C5,10 10uF, 25V tantalum
SEMICONDUCTORS
BR1,2 $1 \mathrm{~A}^{4} 200 \mathrm{~V}$
IC1,2 LM 723CN
Q1,2 TIP 31A
D1,2 2 Red LED
MISCELLANEOUS
$\begin{array}{ll}\text { T1 } & 0-17 \mathrm{~V} ;, 0-17 \mathrm{V5} \\ & \text { secondaries; } 20 \mathrm{VA} \\ \text { transformer. }\end{array}$
$\mathrm{F}, 2$ : 1 A fuses with PCB holders.
S1 DPST rockerswitch.
Hoatsink ( $4^{\circ} \mathrm{C} / \mathrm{W}$ ); mains connector; output socket.
The above will fit into a Teko Alba A23 case.


Fig. 2 (above). Component overlay for the PSU project.


Left: the inside story on the PSU project. The heatsink is mounted on the rear panel of the case.

The VCO is based on the CEM 3340 IC produced by Curtis Electromusic Specialities and specifically designed for music applications. Figure 3 shows the block diagram of the device which is packaged in standard 16-pin DIL form. It provides the exponential law required for music, that is, a doubling of the frequency for a fixed increment of input control voltage, which for this project has been set at one volt per octave. The design provides excellent conformity to the exponential response from less than 10 Hz to over 16 kHz .

The CEM 3340 has triangle, sawtooth and pulse outputs. In the design presented a sine wave has been generated from the triangle output. The outputs are 0 to 10 V in amplitude which facilitates their use as control signals. Additionally, the triangle and sine waveforms are provided at $\pm 5 \mathrm{~V}$ amplitude and are directly suitable for modulation applications. The duty cycle of the pulse output can be varied from 0 to $100 \%$, either manually or by external voltage control. A linear frequency modulation input with a $10 \%$ change in frequency per volt is also included.

The IC allows three methods of synchronising the oscillator which provides an exceptional range of modulation and harmonic locking effects.

The circuit is shown in Figure 4.

## Construction

Construction is straightforward and the component layout on the printed circuit board is shown in Figure. It is essential to use the components specified if the accuracy and stability of the design is to be realised. All jack sockets and potentiometers should be wired up except for the wiper of the fine control (RV2). Neat and short wiring reduces the likelihood of crosstalk and all inputs and outputs are provided at the front edge of the PCB to avoid excessive wire length. The PCB will fit the mounting lugs in a Teko Alba A23 case. It may also be mounted to the $228 \times 76 \mathrm{~mm}$ panel with proprietary brackets or L. brackets. If the latter are used a 12.5 mm spacer between the panel and the bracket is recommended since this provides adequate space for the wires.

When all components have been installed carefully check their placement and orientation and also inspect the underside of the board to ensure that no solder bridges have been made. Before

## MODULE 2: VCO




Fig. 3. Block diagram of CEM 3340 VCO chip.
inserting the IC it is good practice to connect the board to the power supply and check that power is available at the correct points in the circuit and this is simplified by using IC sockets. If this step is satisfactory then disconnect power before inserting the IC.

A final check before calibration is to ensure that all outputs are functioning. Connect the $0-10 \mathrm{~V}$ triangle to an oscilloscope or amplifier (turn volume control nearly off) and then power up the circuit. If there is no response then switch off immediately and repeat the checking procedure. If functioning then check the other outputs although remember that the sine waveform has not been trimmed at this time.

## Calibration

Set all trimmers to their mid positon.
The first step is to adjust the sinewave output. With an oscilloscope, or by ear via an amplifier, adjust TP 4 then TP 5 for purest sine output. These adjustments may have to be repeated a few times to obtain the best results. Next adjust TP 6 to get the 10 V output referenced to ground. The simplest method is using an oscilloscope, but it can be trimmed using another VCO. If the latter technique is used, set the other VCO to about 1 kHz and the frequency of the VCO being calibrated at a low frequency (coarse control switch on and RV1 fully anti-clockwise). The sinewave output is then plugged into the other VCO and TP 6 adjusted until there is no discernible jump in frequency.

The last and most important step is to calibrate the oscillator to the IV/ octave relationship. The easiest method requires a variable voltage source (from a potentiometer or a calibrated keyboard), an accurate voltmeter and a digital frequency meter.

Alternatively, if a previously calibrated oscillator is available then the beat frequency technique may be employed. Another approach is to build two stable fixed frequency oscillators and use these in conjunction with an oscilloscope to calibrate the VCO by generating Lissajous figures. Whichever method is chosen the calibration proceeds as follows. With the coarse control switch, SW1 in the OFF position apply a positive voltage to Control Input 1 (R5) until the frequency is about 200 Hz .

Increase voltage by one volt (as accurately as the measuring equipment allows) and adjust TP 1 until the frequency is doubled. This step is repeated until a doubling of frequency for a one volt change in input is achieved. Next increase the frequency to about 5 kHz , increase voltage by one volt and adjust TP 3 to obtain doubling of frequency. Finally, re-check the first step and if a DFM is available check the VCO over its full range.

Note that the oscillator has been (accurately) calibrated for Control Input 1. This input should be used for the keyboard, if fitted, since there may be a slight difference between Control Inputs 1 and 2 even though 1\% resistors are used.

The oscillator may now be adjusted so that with no input voltages it will be tuned to the lowest frequency of a four
octave C-C keyboard. Adjust TP 2 to 65.4 Hz , or if connected to a keyboard press note $A$ and adjust to 440 Hz . This step is not essential until the VCO is connected to a keyboard. Also R4 and TP 2 may be changed to suit other keyboard requrements.

The wire to the wiper of the fine control potentiometer (RV2) may now be connected.

## HOW IT WORKS - VCO

## Voltage Controlled Ocollitor

The VCO utifiles ar integrated circuit which has been mectically devenal for muadc applications Understaniting the $1 C$ and how it overcomes the many problems ats ciatied with building expenentic cenvertern from theref components y best epproached by consfdering the latter ctrcuitio. A transistor dosign makes nee of the feet that in an ordinury transistor the collector. currentix is an exponential function of the baseemitter voltare end the current, can then be used for say, charging a cappecitor. The omitty satuetign current, however, doubles for every $10^{\circ} \mathrm{C}$ change in temperature and to to employ:a single transistoy necessitater use of téchnlquer such as 'ovening' . that is, matnituning tho transiltor at an olevated temperstuis. Al better approach is to use a well thatched transistor pair, , preforebly pri the ame chip, and to arrange them so as to cancel out this partieu. lar temperature dependent term. A simplitiod osclllator input stage using two transistors if shown li Figure A below.

iCl and associated rexistor and trimmer pro vide a means of scaling and ranging the input voltages and for The standard it V/octave response the gomponenttive chosen to give approximately 18 mV increase at the base of Q 1 for an increase of I V at Vini Noxt tho control range is adjusted by applying arrefer ence current to the collector of Q I. Xdeally this would be in the middle of the exponential range of the transistor, about luh, but considerations of bias curforit from 1 C 2 usually dictate that Ints is 10 u A, or mare, IC2 also serves to sink the excess currentfrom the emitters of Q 1 and Q 2. Thus the eurrent from the output transistor Q 2 becomes:-

$$
I_{\text {out }}=\mathrm{ref}^{\mathrm{V}} \mathrm{~g}^{\mathrm{q} / \mathrm{kT}}
$$

so that with $\mathbf{V}_{16}=0$, Joul =1ret. There te mains, however, a temperature dependent ( $\mathrm{ermm}_{4}, 1 / \mathrm{T}$ which changes the exponent by
sbout 0.33\% for wery degree chativite chang in tomperther To compion tr for

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This groject it teted on the CEM 330 Voltege Controllod Quellt ton produced. 6 Curtil Electromialc spockithe Reforence to the timptiliod block absum ( Figure 3) show thet the whole of the linput stige described above hat been hicospioited within this le Py is is the impat comming anct whto Pin 14. Terermbies the scate. Smoe the curfent gain of the multultas is net chite uilty a 100 k
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 of the oscilitor us toctaves switi 81 he been fited betpeans VI and R7 so that normal use; he, theyberef prectse octaje shift; and extemal voltege coatrols, ilight yariation in RVF will not cause hio oscllitor to go out
 ment of approximatidy 10 S octaves. The other components (RO, C5) on the summint input are for componeation and are ywiyi required. The sum of the fipat volzages should always remain portive for proper operation of the oscilitor.
For greatest multipher werney the chim rent flowing out of Pin 1 (20)VT/R2, where $V T=26 \mathrm{mV}$ at $20^{\circ} \mathrm{C}$ ) thot (be close to the current flowing out of Pe in (3.0/R1 + PP1). TP 1 is used to adjust the opcillator to a pre-- 1 V/octuye response.

The exponential rientor is capable of dellivering a current the tor charging and discharging the tming capacitot; from getater than 500 vA, to low than the inpit bla curent of the buffor which tives typical frequency range of $500,000:$. For typuthe siser applications one should nee the thost accurnte portion of this rainge'; which is from 50 nA to 100 uA . Theso current kinits aro used to tetermine the value of the timing sapacitor, Cr (C7). The oscillation frequency is given by fo $-31 \mathrm{led} / 2 \mathrm{VcoCT}$ and therefore to keep within the above cuncont limits and to maintain an accurate frequency range of 5 Hz to 10 kHz the capacitor will be 1000 p with a Vce of +15 y
An Irefi of 15 eq is produced by R11 connected to the tis $V$ supply: This haput (Pin 13) is also used for linear frequency modulation of the oscillator R12 adjusts the RM
range to a $10 \%$ change in frequency per volt and an Lenuating potentiometor (RV3) alow minual adjustment of this range. The PM loput has boon AC coupled to sa to avoid errory from any DC dift although it may be DC coupplod if required. A negative voltage at Pin 13 will gate the oscillator off.

Reference was made earlier o the need to provide temperiture compensation. One of the many novel features of the CEM 3340 is the incorporation of temperature compensation within the device. This is achieved by multifitying the current sourced into the cons tral Pin (Pin 15) by a coefficient directly proportional to the absolute temperature This coefficient is produced by the Tempco Generator' using the same mechanism as in the uxponential generator and thus cancellation is thearly perfect.

4 furthor problem that occuir with frandator exponential converters is their buik emiterer resiatance which becomes a signfficant fletip as current is increased and will cause the osctilitor to go flat With the CEM 3340 this thention applies when current from $\mathrm{C}^{2}$ Wrimeter than $50 \% \mathrm{~A}$. Means of correcting Wor thin offect have been included since Fin 7 outputs if current which is a quarter of the exponentha generato curient. The current is converted to a voltage across TP 3 and a proportion cin be fed back into the control input Ie R10.
Waveform Outpintic All wayeform outputs trom the IC are shibtt circuit protected fand may bo shorted continuously to any supply without damaging the devico. A $0<10 \mathrm{~V}$ sawtooth waveform is available at Pin 8 which can stak at least 0.6 mA and sourco over pyeral miliamps withoul any effect on osciHotot performance and only a negigible effect on waveishape. The pulse oufput from Pin 4 is an open NPN emitter and therefore requires - pull downetosistor to ground of a degative voltage. Thissoutput has been clamped with $x$ 10 V zeher diode (ZD 1) to give a $0-10 \mathrm{~W}$ pulse output. Pin 5 allows pulse width moduhaton and 0 to 5 V applied to this pin will viry the pulse-width from 0 to $100 \%$. Attenuating resistors R19, R20 increase the control renge to orer standard 0 to 10 V. P4 connected to $+15 \times$ provides manual control of pulse width and this control voltage is further attenuated by R18. RV4 input is connocted via a jack socket so that it is disabled when an external voltage control source is used. Pin 10 outputs a $0-5 \mathrm{~V}$-triangle waveform. Although the sink and source capabilitios approach those of the sawtooth this output has 's inite impedance and also drives the' comparator with the result that loading of this output into 100 k impedance may lower frequency by $0,15 \%$ in the worst case. This


Fig. 4. The circuit diagram for the Voltage Controlled Oscillator block used in the project 80. Note that with some component value changes, this is the same configuration as used in the low frequency VCO (module 3).

output has therefore been buffered of lC2 and tudition of R14, R15 increases grim bes two to give $10-10 \mathrm{~V}$ trangle output

The triangle output is converted to tst at IC 2 b (this output is made available freas the $V(0)$ and attenuated to about $\pm 100 \mathrm{mV}$ by TP 4, R 30 and R31 prior to IC3 which 45 a CA 3080 E . Use is made of the non linear characteristics of the OTA at high input lovols to convert the triangle into a sine wave. 78 adjusts the 3 rd harmonic content and TF 5 and associated components at the inverting saput of IC3 trim the 2nd harmonic. IChe converts the current from IC3 to a voltage and provides a $\pm 5 \mathrm{~V}$ sinewave. This output is $180^{\circ}$ out of phase compared to all other outputs but this is not detrimental in most appli. cations of this output. Finally the sine wave fo shifted to a $0-10 \mathrm{~V}$ imput by IC4b with R 37 and TP 6 as the level shifter.
Power supply to IC: The CEM 3340 will oper ate directly from positive supplies of between 10 and 18 V (although the amplituds of the waveforms as given above will decrease and increase respectively) and negative supplies


 eartiots



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 grachioniong oftiots than foand of wom wh
 chronimation by nagitive pulites se tif 9 caues thertrimple uppet poth th, whets direction promaturoly wtin the rault thit the ovcillation perfod if an interral multhpe of tít. pulse period. If this mpat will not bo wel Chould be byyinva to gound vith 1 too
 the PCD). Adtematively the 100 n orpactor



Fig. 8.


## PARTS LIST - VCO



TRIMMERS

| TP 1 | 10k cermet multiturn |
| :--- | :--- |
| TP 2 | 100 k cermet |
| TP 3 | 10 cermet multifurn |
| TP 4,6 | 10 k carbon |
| TP 5 | 1 M carbon |

POTENTIOMETERS
RVI, 2,3,4 100k lin.

## SEMICONDUCTORS

IC1 CEM 3340
IC2 LM 1458
IC3 CA 3080E
IC4 TL 082
D1,2 IN4148
ZDI BZY 8810V

## Miscellaneous

Si SPDT miniature switch; 3.5 mm jack sockets ( 13 )


Fig. 5 (above). Component overlay for the VCO/VCLFO modules. The parts list to the left is for a standard VCO circuit. Use the changes given to the far right of page facing to convert to a low frequency oscillator.
Below: A made up PCB sat sitting in its box.


The VCLFO is identical in design to the VCO described above. It is designed to operate in the range of 0.2 to 205 Hz with a 10 V control input. It is, however, capable of about 5 kHz . Alternatively a lower minimum frequency may be obtained by using a higher value timing capacitor, C7, provided that the capacitor is a very low leakage type.

Note the component changes listed. Construction and initial setting up follows the same procedure as the VCO except that the fine control potentiometer can be wired up prior to calibration.

## Calibration

Set all trimmers to their mid position. Adjust sinewave output as described for VCO.
Turn both coarse and fine controls fully anti-clockwise (zero input) and apply a voltage to Control Input 1 to obtain doubling 100 Hz . Increase voltage by one volt and adjust TP 1 to obtain doubling of initial frequency. Repeat step until the one volt per octave response is achieved. Next increase frequency to about 1500 Hz , increase voltage by one volt and adjust TP 3 to obtain doubling of frequency. After adjusting TP 3 re-check setting of TP 1. Finally with 10 V applied to Control Input 1 adjust TP 2 to give a frequency of 205 Hz which will result in a frequency of 0.2 Hz with no external input voltages applied.

ET

## BUYLINES

The Dowar supply PCB and compoTontw. includiti'g the heatsink and Hmenheted rocker switch but not the Wheran ave allable from Digitwen for $\mathbf{~ 1 9 : 5 5 ~ i n c l . ~ p o s t a g e ~ a n d ~}$ WHT:
The Pal and components lincluding 4-2) To each voltage controllea 4- tititor cre wallsble fromi Digisound for E1 6.63 inel. postages and VAT.

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ge The trooklands
STia Green
Fruston
Lenceantre PR4 2 NO
The modules are case in Feko Alba
A 286 ceses, evailable from
What Hyde Developments Ltd
thit9
Werk Street Industrial Estate
Aytelbury
Bocla HP2O IET
Thay roet C4:43 each incl. postage and
VAT, Plesse use order code TEK
A23G.

## MODULE 3: VCLFO



Above and below: the inside and outsides of a low frequency VCO. The PCB is panel mounted just to show we did both ways! This method of construction is best suited to professional (or very serious amateur) use.


## HOW IT WORKS - VCLFO

Voltage Controlled Low Frequency Osciliator The LFO is identical in design to the VCO described above. A larger timing capacitor (C7) is used to give a frequency tange of between 0.2 and 200 Hz with a 0 to $10 . \mathrm{V}$ control voltage. R4 and TP 4 are used to set the lower limit; RV2 provides an adjustment, of one octave; and RV1 a range of 10 octaves. A switch is not required betweên RVI and R7 in this instance. One reason is that both $\mathrm{RV}_{1}$ and RV2 are connected between +15 V and ground and so when the controls are fully anti-clockwise the oscillator is at the preset frequency ( $0: 2 \mathrm{~Hz}$ ). Secondly minor varb ations in frequency arising from RV1 and RV2 are of little importance for tow fre quency applications.

## PARTS LIST <br> - VCLFO

## COMPONENTS AND SYSTEMS FROM TRANSAM COMPUTERS



## DPS. 1 MAINFRAME - PASCAL SYSTEM S 100

 Tol IEEE SPEG $\qquad$ S100 BOARDS8k Static RAM board (450ns) $\mathbb{1} 123.75$
8k Static RAM board
(250ns)
$\begin{array}{ll}8 \mathrm{k} \text { Staric RAM board (250ns) } & \text { E146.25 } \\ \text { 280 cpu board ( } 2 \mathrm{MHz} \text { ) } & \text { E131.25 }\end{array}$
$\begin{array}{ll}280 \mathrm{cpu} \text { board ( } 2 \mathrm{MHz}^{2} \text { ) } & \text { £131.25 } \\ 280 \mathrm{cpu} \text { boord ( } 4 \mathrm{MHz} \text { ) } & \text { £153.75 }\end{array}$
$2708 / 2716$ EPROM board $\begin{aligned} \text { E83.75 }\end{aligned}$
Protorype boord (bare board)
Video display board ( $64 \times 16$
128 l
Disk contuoller boara a
$\mathbf{F} 108.75$
$\mathbf{f} 31.25$
K 2 dish operating system $\quad$ E53.25
ASSEMBLE/Z Macro Assm
Pascale $/ 2$ compitet
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| :---: | :---: |
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| CP249 $1 \times 51 / 4 \mathrm{PSU}$ | ¢33.00 |
| CP323 $2 \times 51 / 4 \mathrm{PSU}$ | ¢80.00 |
| CP205 $1 \times 8^{\prime \prime}$ PSU | ¢56.00 |
| CP206 $2 \times 8^{\prime \prime}$ PSU | ¢76.00 |

TCL PASCAL CP/M compatible. A standard Pascal compiler available ONR resident (20k) Eprom based con figuration or available to run under $C P / M$ on $8^{\prime \prime}$ disk plus documentation. $\mathrm{CP} / \mathrm{M}$ version $\mathbf{£ 9 0}-+$ P.O.A.

## DIL PLUG SOCKETS \& SWITCHES

| W/wine sits |  | DM sars |  | on fucs <br> 1481 |  | 0am swicmes |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
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| me | 230 | ${ }^{6046}$ | 0.14 |  | 0.6 | 70 m | 1.55 |
| 1401 | 0.35 | 14010 | 0.15 |  |  |  | 8 ロa | 190) |
| 160 | 0.42 | 16001 | 0.17 | SCUTCH Re |  |  |  |
| 1800 | 0.90 | 1801 L | 0.24 | 14 DL | 230 |  |  |
| 2401. | 0.52 | 2000 | 0.27 | 16 DM | 4.50 | 16 maF | 4.58 |
| 2800 | 6.74 | 240 FL | 0.30 | 24 吅 | 20 | 24w 3F" |  |
| 4000 | 0.5 | 2800 | 0.30 |  |  |  |  |
|  |  | 4800 | 0.90 |  | MSER | F FOREE. |  |

COMPUCOLOR II - FULL COLOUR



## CATALOGUE

 1980

## Gentroninstotary <br> international

## What to look for in the March issue: on sale February lst



## TV SOUNDS GOOD?

Tired of tinny tunes from your telly? The melodic meanderings start out from the transmitter in super-duper hi-fi, but the cost cutting sounds section of your set takes care of that, lowering the fi at the speed of light. Next month Richard Maybury explores the world of TV sound and comes up with a few ideas on improving it.

## THE ULTIMATE METAL LOCATOR

Calling all treasure hunters. How many times has your metal detector gone ping or buzz or hello sailor and you've shifted half a ton of Surrey only to find a non-biodegradable ring pull tab? Well, next month we have a discriminating metal locator for you.

The magic machine rejects nails, bottle caps, aluminium foil and ring pull tabs. The design also features full ground effect exclusion over normal or high permeability soils. Search for your pot of gold with deepseeking VLF plus three TR discriminating ranges. Instant tuning recall is made possible by a push button memory circuit.

## BLACK HOLES

When a massive star reaches the end of its life, uses the last of its nuclear fuel and explodes as a supernova, one of three things can happen. The supernova explosion may destroy the core, or, if a small core remains, it may become a neutron star, or, if it is large enough, it may collapse to form a black hole.

Next month Ian Graham has a bash at explaining that most enigmatic of astronomical propositions - the black hole.

## HEATER POWER CONTROLLER

With most heater controllers, your heater is either on or off and the room temperature fluctuates several degrees either side of 'comfy'. Our design will keep your room temperature stable to within half a degree. In addition, by using zero voltage switching, RF interference is avoided.

If you're into Biofeedback you can use the ETI Muscle Meter to learn to relax more effectively. On the other hand, if you're into having fun, there's plenty of scope for doing your own thing. Watch this space (give or take a few pages) to find out how the miracle machine picks out the fractions of a microvolt of relaxed muscles from the volts of 50 Hz hum present in the body - induced from power and light wiring.

## ELECTROMYOGRAM

The ETI Muscle Meter senses the tiny electrical impulses associated with muscle activity. As Superman flexes his biceps you can hear it all happening and see the activity building up on a meter.

ETI576 ELECTROMYOGRAM
 MUSCLE ACTIVITY

## Project editor Ray Marston makes an in-depth review of the revolutionary mini-robot, HEBOT.

The basic electronic concept of the Hobby Electronics HEBOT is brilliantly innovative and uses the priority encoding principle illustrated in Fig 1. Each drive motor is energised fron an analogue driver circuit. Each driver has a high input impedance and can drive its motor forward or backward at a speed proportional to the input voltage. The motor drive 'instructions' 'analogue voltages) can be selected from any one of (up to) eight independent sources (or circuit boards) via a motor instruction selector or 'multi-way electronic switch', which has its position controlled by an (up to) eight-level priority encoder. This encoder causes the motor to take its instructions from the source or circuit board with the highest active priority coding at any given moment of time and is analogous to a human central nervous system.

Suppose that we connect just four independent circuit boards to the HEBOT's central nervous system and assign each board a priority number as shown in Fig 1. Board 1, with the lowest coding, simply instructs the motors to drive HEBOT in a 'search' pattern. Board 2 is a light-sensing
circuit that gives a 'high' output to the priority encoder only in the presence of bright light and then produces motor instructions that steer the HEBOT towards the light source.

Board 3 is designed to detect the presence of, and guide the motors along, an inductively-coded wire loop. This board produces a 'high' priority output only in the presence of the loop. Finally, board 4 jwith the highest priority coding) handles the tactile sensor logic and produces a 'high' priority output in the event of a collision and then instructs the motors to execute a suitable avoidance manoeuvre.

## A Sequence Of Actions

A possible sequence of HEBOT actions may in this case be as follows.

At switch-on, all priority outputs except ' 1 ' may be low, so HEBOT accepts instructions from board 1 and goes into a 'search' or rotary-movement mode. During this move-


Fig 1. The priority encoding principle used in HEBOT has revolutionary implications in Robotics.
ment the light-sensing board may detect a powerful light source some distance away and so set its priority output high and start generating the motor instructions necessary to steer the HEBOT towards the light source. Since ' 2 ' is a higher priority number than ' 1 ', the priority encoder will automatically turn the motor instruction selectors to board 2 and HEBOT will start moving towards the light source. Note that under this condition the 'search mode' instructions may still be generated but will be ignored by HEBOT.

Suppose now that, as HEBOT moves towards the light source, board 3 detects the presence of an inductively coded wire loop. In this case the priority- 3 output of the board will go high and HEBOT will lock on to the motor drive instructions of board 3 , ignoring the instructions of boards 1 and 2. While following the inductive wire, HEBOT may collide with an object. The collision will be detected by board 4 the highest priority board) and the mini-robot will then execute an avoidance manoeuvre based on instructions from board 4, ignoring all instructions from boards with lower priority numbers.

On completion of the avoidance manoeuvre, HEBOT may have lost contact with both the inductive wire (priority 3 ) and the light source (priority 2), in which it will revert to the priority- 1 'search' mode until a higher priority number is generated.

## A Revolutionary Concept

This 'priority coding' system has enormously important and revolutionary implications in robotics. First, it means that robots no longer have to be designed as single complex entities, but can be built as highly versatile devices containing a number of quite independent control/sensing modules, all feeding ino a simple central nervous system or priority encoder.

The simple (?) version of HEBOT published in the recent editions of Hobby Electronics uses only the four priority levels already described. The experimenter can, however,

| 8 | RADIO <br> CONTROL |
| :---: | :---: |
| 7 | COLLISION AVOIDANCE, <br> (TACTILE) |
| 6 | COLLISION AVOIDANCE, <br> (ULTRA-SONIC) |
| 5 | HUNGER AVOIDANCE |
| 4 | WIRE FOLLOWER MODE <br> [SEARCH AND LOCK] |
| 3 | OPTICAL RESPONSE |
| 2 | ACOUSTIC RESPONSE |
| 1 | 'SLEEP' MODE |

Fig 2. A possible sequence of priority coding for an advanced version of HEBOT.
design additional control boards and add them to the HEBOT at will. Figure 2 shows a possible way of connecting a number of independent circuits into the HEBOT central nervous system to give eight levels of priority encoding.

Individual boards can even incorporate their own priority-coded response conditioners, in which case they can be regarded as NERVOUS SUB-SYSTEMS

## Human Behaviour

The second important implication of the HEBOT priority coding system is that it can give a very close analogy of animal/human behaviour. Figure 3 is a provisional table of human behavioural priority levels. The table shows six main priority levels, with COMFORT/PLEASURE ;into which work ethic is entwined) having the lowest priority level and CONFORMANCE (conditioned resposes) having the highest priority level (even higher than survival). If a

| 6 | CONFORMANCE <br> (CONDITIONED RESPONSES) |
| :---: | :---: |
| 5 | SURVIVAL |
| 4 | PAIN AVOIDANCE |
| 3 | SLEEP <br> (RECOUPERATION) |
| 2 | HUNGER <br> SATISFACTION |
| 1 | COMFORT <br> 1 <br> WORK <br> ETHIC <br> PLEASURE |


| 4 |  |
| :---: | :---: |
| 7 | DESTRUCTION <br> AVOIDANCE |
| 6 | EXHAUSTION <br> AVOIDANCE |
| 5 | STARVATION <br> AVOIDANCE |
| 4 | DAMAGE <br> AVOIDANCE |
| 3 | ATTACK <br> AVOIDANCE |
| 1 | THREAT <br> AVOIDANCE |
| SPPECIES <br> SURVIVAL |  |

SURVIVAL PRIORITY SUB-LEVELS

Fig 3. Provisional chart of human behavioural priority levels and survival sub-levels. Behavioural patterns are predictable and can be simulated by a HEBOT-type robot.


PLAN VIEW


A = MOTOR DRIVE UNIT<br>B = BALL CASTOR<br>C = TACTILE SENSOR

number of these priority levels are stimulated simultaneously, the highest priority will override the lower ones.

The really fascinating feature of human/animal behaviour is that each of the main priority levels consists of a number of sub-levels. In Fig 3 seven main sub-levels of the survival priority are shown. Each of these sub-levels can in turn contain priority sub-levels. Each level and sub-level is fed with sensor information which itself has a number of basic levels: peckishness, hunger and starvation are, for example, three levels of the same basic sense which, when connected to the appropriate levels and sub-levels, produce highly complex but predictable behavioural patterns.

Similarly, work ethic normally has a low rating in the main priority levels but may, under certain circumstances, attain very high priority ratings as discrete sub-levels of a survival (level 5) or conformance (level 6) main priority level.

These complex behavioural patterns can easily be electronically simulated, or modified and improved upon. by using HEBOT's basic and revolutionary 'priority encoding' concept.

Fig. 4. (Above) The HEBOT chassis. Although outwardly very simple, a great deal of thought and effort went into the final design of this. Anyone who saw HEBOT trotting around at Broadboard exhibition recently will realise that this configuration is in no way a limiting factor on HEBOTS performance.
The shape of the chassis helped us to determine several important factors for HEBOT and helped in finalising the tactile senser arrangements.
highly sophisticated super-robot. At the other end of the scale, it is feasible that a highly sophisticated HEBOT-type robot could (like a living animal) have $90 \%$ of its circuitry blasted to shreds by machine-gun fire yet still be able to perform useful function (stay alive). in this latter case, the 'vital organs' of the HEBOT are the central nervous system, the motor drive system and the power supply unit.

You can see, then, why we are so proud of HEBOT. Superficially, it does not look very impressive, but it in fact incorporates the most far reaching design concept yet devised in the field of robotics. It represents a major breakthrough in robot technology and places the off-theshelf commercial/domestic robot of 'science fiction' within the reach of all of us. And, by jove, it's British!

En

## Indestructible

The final important implication of the HEBOT 'priority encoding system is that, since most priority levels and sub-levels are normally low or inactive,, it follows that total obliteration of an individual priority level will resull in a mere degradation of the total performance of, rather than the total disabling of, a priority encoded robot system. Compare this concept with that of a television or computer. where every individual component is vital to the functioning of the total system! An alternative way of stating the same point is to say that the performance sophistication of a priority-encoded robot can be enhanced at any time by merely adding extra priority levels or sub-levels to its existing 'nervous system'.

Thus, at one end of the complexity scale, our simple little HEBOT project can (in principle) easily be expanded into a


| CMOS |  | 4020 | 50p | 4050 | 25p |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 4022 | 13.) | 40 | \% |
|  |  | 4024 | 40, | 4068 | 30 |
| 4001 | 13p | 4025 | 13, | 406 | p |
| 4002 | 13p | 4026 | 901 | 4070 | 130 |
| 4007 | 130 | 402 | 280 | 40 | 130 |
|  | 30 | 40 | 45 | 40 |  |
|  | 1 | 402 |  |  |  |
| 4012 | 13p | 4040 | 551) | 409 |  |
| 4013 | 28 | 4041 | 55p | 451 | 60p |
|  | 50, | 4042 | 55p | 451 | 600 |
| 4016 | 28 | 4043 | 50 | 45 |  |
| 4017 | 470 | 4046 | 901 |  |  |
| 401 | 55p | 4049 | 25 | 45 |  |

## FULL DETAILS IN CATALOGUE!

## 

LED's 0.125in. 0.2.n each 100, Red TIL209 TIL220 9r 75p Green TIL211 TIL221 13p 12p Yellow TlL213 TLL223 13p 12p Clips 3p
DISPLAYS
DL704 $0.3 \mathrm{incC} \quad 130 \mathrm{p}$ 120p
$\begin{array}{llll}\text { DLL707 } & 0.3 \mathrm{hnCA} & 130 \mathrm{CA} & 1200 \\ \text { FNO500 } & 0.5 \mathrm{nCC}\end{array}$

## SKTS <br> Low profile


 3 lead TO18 or TO5 socket. 10 p each Soldercan pins: $100: 50 \mathrm{p} \quad 1000: 370 \mathrm{p}$

## PCBS

veroboard
$\begin{array}{lll}\text { Size in. } & 0.1 \mathrm{in} 0.015 \mathrm{in} \text {. Vero } \\ 2.5 \times 1 & 14 p & - \\ \text { Cutter } 80 p\end{array}$ $\begin{array}{llll}2.5 \times 1 & 14 \mathrm{p} & - & \text { Cutter } 80 p \\ 2.5 \times 3.75 & 45 \mathrm{p} & 45 \mathrm{p} & \end{array}$ $\begin{array}{llll}2.5 \times 5 & 54 \mathrm{p} & 54 p & \text { Pininsertion } \\ 3.75 \times 5 & 64 \mathrm{p} & 64 \mathrm{p} & 10\end{array}$ $3.75 \times 5 \quad 64 \mathrm{p} \quad 64 p$ tool 1080
Single sided
pins per 100 40p 40p
Top qualitv fibre glass copper board Single sided. Size $203 \times 95 \mathrm{~mm}$. 60p each
'Dalo' pens. 75 each

| RESISTORS |  |
| :---: | :---: |
| E12 seres 4.7 ohms |  |
| 促 | 100+ |
| $1.5 p \quad 1.28$ |  |
|  |  |
| (e) |  |
| Very high stability, low noise rated at $1 / 4 \mathrm{~W}$ |  |
|  |  |
|  |  |
|  | $100+$ 350 |
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| OF Compon |  |
| ENTS. |  |
| VER 2 |  |
| ITEMS LISTED |  |



## CAPACITORS

TANTALUM BEAD
$0.1,0.15,0.22,033,0.47,0.68$,
$1 \& 2.2 \mathrm{uF}$ @ 35 V
$22 @ 16 \mathrm{~V}, 47 @ 6 \mathrm{~V}, 100 @ 3 \mathrm{~V}$
MYLAR FILM
$0001,0.01,0022,0.033,0.047$
POLYESTER
Mullard C280 serie
$0.01,0.015,0.022,0.033,0.047,0.068,0.1 .5 p$ 0.15 .0 .22
0.68
$5 p$
$7 p$
0.68

CERAMIC
Plate type 50 V . Available in E12 series from 22 pF to 1000 pF and E 6 series from 1500 pF to 0.047 HF

RADIAL LEAD ELECTROLYTIC
$\begin{array}{lllllll}63 V & 047 & 1.0 & 22 & 47 & 10 & 5 p\end{array}$

|  |  |  | 22 | 33 | 7 p |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 100 |  |  |  |  | 13p |
|  |  | 220 |  |  | 20p |
| 25 V | 10 | 22 | 33 | 47 | 5p |
|  | 100 |  |  |  | 8 p |
| 220 |  |  |  |  | 10 p |
|  |  |  |  |  | 470 | 15p |
| 1000 |  |  |  |  | 23p |

## CONNECTORS

JACK PLUGS AND SOCKETS

|  | unscreened | screened | socket |
| :---: | :---: | :---: | :---: |
| 2.5 mm | 9 p | 13p | 7p |
| 3.5 mm | 9 p | $14 p$ | 8 p |
| Standard | 16p | 30 p | 15p |
| Stereo | 23p | 36p | 18p |
| DIN PLUGS AND SOCKETS |  |  |  |
|  | plug | chass is | line |
|  |  | socket | socket |
| 2pin | 7 p | $7 p$ | 7 p |
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# ¿ THIS YEAR'S STAR BUY $\nsim$ 

 THE NEW 830S - 27B ALARM CHRONOGRAPHOptional display of hours, minutes, seconds, date, am/pm; or alternatively: Hours, minutes, alpha day, date, $a \mathrm{~m} / \mathrm{pm}$. The automatic calendar is set for 28 days in February. Casio's new Lithium battery lasts up to 4 YEARS or more. The chronograph times in $1 / 10$ second units up to 12 hours, measuring net, lap and first and second place times. An indicator shows the chronograph is running when normal time is displayed. The 24 hour alarm can be set very easily to 1 minute intervals with an indicator to show the alarm is set.
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F-8C Time/date 3 YEAR BATTERY Hours, minutes, seconds, datt, day, am/pm. Aulo 2B, 30,31 day calendar. Backlight, Resin case strap. Mineral glass. W.R. to 66 feet.

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## TUNING FORK

# You don't have to bash this tuning fork on its head and sit it on the sideboard to get a note. Just flick a switch and get a constant frequency tuning tone. 

Musical instruments are generally a clever and harmonious blend of physics and aesthetics (both aural and visual). But, underlying all this wonderful harmony and human cleverness is the law of 'the cussedness of nature'. This law, simply explained, says that through all the consistency and harmony we find in nature runs a streak of cussedness always causing something to be out of place. It is this very streak of cussedness that has thwarted attempts to date to develop a 'unified field' theory that would link gravity, electricity and magnetism.

It doesn't seem, at this stage, that gravity, electricity and magnetism have much to do with musical instruments and tuning forks, but we'll get around to it!

For one musical instrument to be played with another requires both to be tuned to the same fundamental pitch (or frequency). If not, the sound will be unpleasant - generally described as discordant.

## Fork Lore

Over the centuries there were various ideas as to what basic 'standard' pitch would be adopted. After some considerable squabbling a 'standard concert pitch' was settled upon in 1929. This gave the note ' $A$ ' a pitch of 440 Hz . That standard remains to this day. Of course, it means that modern orchestras playing the music of Hadyn, Mozart and Bach, for example, will not be playing in the pitch in which the music was originally composed.

The traditional tuning fork consists of two cantilevered bars attached to a common base - it resembles that common eating utensil, hence the name. When the tines are struck (or one tine) they will vibrate, producing a sound of
 pitch ture, except by gross variations, and accuracy can be maintained within about $0.1 \%$.

They are portable and relatively inexpensive but suffer from low sound
level output and do not give a sustained note - it 'dies away'. What's more, as many modern groups use electrically amplified instruments and sound reinforcement, a failing of tuning forks is lack of a pick-up.


Fig. 1. Circuit diagram.


When you get all your bits bolted down into box, it should look like this.

## HOW IT WORKS


#### Abstract

The signal is generated at a high frequency (about 3.6 MHz ) by a cryatal oscillator and then divided down to the output frequericy by a counting circuit. ICle is the osciltion - gates bissod into thetr linear tegion byillis and R2. Capacitor C1 form's \& phases⿱inite network with the bias components, peo viding a shift of 180 degrees at thit crystal frequency. As the crystal is in series witith the feedbeck path, the circuit will oscillete. at the crystal frequency.

LCId forme a buffer botween the oncil lator and the clock input of IC2, a 14 -taje counter.

As the required division is not 2 power of two, decoding of the counter (IC2) out puts is necessary. This is provided in thriee gates - IC1a, IC3c and IC3d. These modify the ouputs of IC2 to obitain the requited division by resettitig IC2 after the appre priate count

Switch SW2 changes the decoding for either a division by 8128 for a 440 Hz output or 8048 for a $445^{\circ} \mathrm{Hz}$ output.

When all the inputs of $1 \mathbf{C 4}$, in dight input NOR gate, go high its output goes-low and drives CLI via a network to remove noise pulses. (R3,C2), IC1d then plovides a reset signal to the divider, ready for the sext count.

The Q13 output from the divider provides a signal at the required frequency and, after buffering provided by IC3a is fed to the volume control. The pulses are then fed to an emitter follower ( $Q 1$ ) and thence to the speaker.





Again, the cussedness of nature raises its head. Remember too, the popularity of the electric guitar. They have magnetic pickups and require plugging into an amplifier. Now you see what electricity and magnetism have to do with musical instruments! Gravity? Oh, most instruments will go out of tune when dropped from a height!

## Construction

We strongly recommend you use the pc board specified for this project. For a
start, it simplifies construction, and secondly it reduces the possibility of wiring errors. With digital circuitry, bugs created by wiring errors can prove most frustrating to track down - particularly if you haven't had much experience with digital equipment. The project is not a difficult one; if you have had a small amount of experience constructing projects and finding your way around circuits and layout diagrams, then it should not prove too challenging.

It is best to commence construction

PARTS LIST


by assembling the components on the printed board. Leave the ICs till last. Solder the crystal, the BC547 transistor, diode, resistors and capacitors in first. Watch the orientation of the diode, D1. Then do all the links using, say 22 gauge, tinned copper wire. There are six in all. Take care here, and refer to the overlay.

The last thing to do is check that you have the switches wired correctly. Make sure that when you switch from 440 Hz to 445 Hz the output goes a little sharp in pitch.

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## Codespeed IEsernonigs

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# conjowing 

WHAT TO LOOK FOR IN
THE FEBRUARY ISSUE
ON SALE JANUARY 11TH.

## LOGIC EMULATOR

Does your logic lie to you? Do you doubt the truth of your tables? Next month we present a solution to your logic design problems with a software logic emulator for the Nascom. The design caters for the usual range of logic gates and will allow you to try out various input combinations. A must for all who are not up to Mr. Spock's logical capabilities!

We start a new occasional series on connections you can make with your micro. In the first part we show you how to get your Mk14 or Acorn to flash lights

## microlink

 on and off, a vital step for micro-mankind.

There have been evaluation kits and there have been teaching aids. Now there are teaching aids which can double as an evaluation kit! What Texas have produced, however, is a one board education system (TM990/189) which is also an excellent evaluation and to the TMS 9980 16-bit MPU!

A 300 page course comes with the board along with a hell of a lot else. But is it good value? Can it teach machine code programming as promised? Could it really educate cavement and Spurs supporters? Can it make good tea?

Don't miss next month's stunning inside information when CT reveals all on the TM990/189. The Wews of the Norld will envy us.

It has been the most (and longest) awaited home computing system for a considerable time. Everyone we spoke to had heard of it - and knew a great deal about it - but no-one had actually seen it!!!

Well we can put you out of the state of suspense because CT has been examining a Nascom 2 in great detail and will report all that there is to be reported in our next issue.

Whatever you were expecting we can guarantee you that the Nascom is a surprise!

## NASCOM 2



# Ron Harris takes time off returning noisy records for a chat with the record industry and a word about mini headphones with a mighty performance. 

1am appalled. Totally and utterly appalled. Apart from that I'm not even happy. In the past month I have attempted in vain to buy myself some LPs with which to exercise all this hi-fi stuff our hobby is based upon.

I was particularly looking forward to 'Tusk'. After all, it cost a large fortune to produce - digital mixing and all and the music promised to be worth all the effort. Six copies later I still cannot comment on how good the production finally turned out. Six copies and six doses of snap crackle and pop made me give up on the idea completely. Sorry Warner Bros but I do like some music with my surface noise. Silly, I know, but I'm old-fashioned that way.

The new Pink Floyd double LP led me a similar dance, wearing out the carpet in the record shop in the process. No-fi 2 me 0, and so it continued all month. Final score? LPs purchased - five; LPs kept - one; copies returned twelve.

Now admittedly this has been the worst ever month for duff records, but it is not unfortunately actually unique. Most records these days have to go back at least once - the record shop next to our office (Our Price Records - and very helpful they are too, thank God) is thinking of giving me a job. After all I spend more time at the exchange counter than the staff.

Other people l've talked to about this seem to suffer from a similar malaise, so I'm not gonna feel too paranoid just yet. Besides, you meet a nice class of person exchanging LPs. Beats discos anytime.

Trouble is - what is the point of all this ultra-fi that some of us are addicted to, if record companies go on churning out pure unadulterated garbage for us to play on it? Digital LPs may be on the way but in the meanwhile we have got to live with the surfaces upon which our music is recorded. Please please please Mr Record Executive Sir take your quality control a bit more seriously. No wonder record sales continue to fall, so does the quality of the product.

If there is anyone out there involved in the industry, engineers, technicians, producers mixers, or even artists themselves, HELP. It's about time we all stood up and got counted. It seems everyone from us poor consumers back down the line wants something done, but no-one is actually doing it!

Well, having added one more small voice to all those already crying into the wind I'll cease and desist - after a quick wave of the fist in the appropriate direction.

## MDR-3 Headphones

Open headphones seem to have continued to gain popularity steadily in recent years and it is never a surprise to see a new pair released. Some years ago Sony had a pair on the market, entitled DR-15s, which I thought to be excellent value for my money. About three weeks ago 1 was very interested, therefore, to see the new MDR-3 open headphones from the very same Sony.

These units are incredibly small and light, and thus comfortable. At first glance they look just too small to complete. At first listen, however, the universal reaction so far has been dropping of chins onto chests. My first thought was to go scrambling through the accompanying paper to find the price. I had expected somewhere in the region of £30-£40.

It came as a shock to read only $£ 16$ in the blurb.
The MDR-3 is a revolution at that price. The sound quality is truly amazing and will satisfy most types of listener. Rock followers who are made a little uneasy by the slightly 'light' balance can apply bass boost all they want with little or no complaints from the MDR-3

Compared against Koss ESP-10 electrostatics, the transducers are shown to be adding a certain amount of 'warmth' to the sound, especially in the midrange, but this is no bad thing on headphones, 1 feel, and is certainly preferable to a hard, colder approach with no room effects to temper it. The treble is well extended and smooth and showed no tendency to harshness or a brittle nature.

The bass is obviously going to be less plentiful than on the much heavier 'can' type of phone but that diaphragm must be moving a long way indeed to produce bass as well as it does.

I have no complaints to make of the bass overall.

## Cryptic Comments

Fitting the phones to the head is simple enough, the cups are angled back so they naturally lie in the same direction as your ears on your head (statement only valid for human subjects. Aliens can figure it out themselves) and have a certain degree of freedom to move and thus align themselves better


The MDR-3s really are small. We tried several young ladies heads, in an attempt to convey just how small - but each time the headphones got lost! Those earpieces are about the size of a 50p piece. But no-one at ETI had one of those either.

As Sony say in their little leaflet, positioning is vital for best performance, so it is advisable to bend the headband to get the best fit with the pads exactly over the ears. Once set, they'll stay there until the moon falls, if necessary, although I doubt if you'll find an LP that long.

The chord is supposed to be oxygen-free copper - a chemical impossibility I would have thought - and is not of the usual coiled spring variety. This I found an absolute pest, as, oxygen free or not, it is pretty good at getting tangled and twisted around the most unlikely of objects

Cobalt magnets are used to achieve the very high field density required to get the sound out of such a small diaphragm, although the leaflet insists that cobalt is used instead of magnets. The diaphragm itself is 12 um thick (so don't go trying to tap it!) and this very small mass undoubtedly confers upon the MDR-3 their quite incredible transient reponse.

Overall, the headphone market at anywhere from $£ 17$ to around $£ 35$ should take quite a beating from the MDR- 3 if there is any justice in the world at all, and I stand by my
comment last month that you'd be mad to buy a pair of headphones in this price range unless you had at least given the MDR-3 a chance to impress you as much as it impressed us.

## Quick Summary

- Small and lightweight $(40 \mathrm{~g})$
- Open type - do not exclude surroundings totally.
- Good extended treble, warm mid-range, good bass for type.
- Will take bass boost without deterioration of sound
- Impedance 32R, thus needs to be driven hard.
- Cord not usual type - inconvenient.
- Excellent value for money, highly recommended


## Guest Moaner

Help! I'm being invaded by people. Ian Graham, ETI's new Celtic Assistant Editor, has cajouled his way into an Audiophile in order to have a moan about racking systems with no racks. Seems a valid point and besides he bribed me with a picture of Felicity Kendal.

So here you go Mr Graham, take it easy at first lad
Put yourself in my position - you're building up your first decent hi-fi system. You save up your pennies, raid your little brother's piggy bank, collect the deposits on every lemonade bottle in sight and generally economise until friends start counting your ribs through your duffle coat. Unit by unit budget lo-fi gradually gives way to higher-fi than you've ever had

Now, the system looks a bit out of place piled up on the coffee table and, besides, it's beginning to sag a bit under all the watts. What about a rack to put it all in - like the one you saw your very own system in in the shop. A quick phone call brings the answer - no, we only supply a rack when you shell out $£ 400$ or so on a complete system Another call to the UK importer brings the same response.

How much nicer and dust-free my Pioneer system would be in the smart, glass-fronted rack where I first saw it. I wonder how many other hi-fiers there are, who must now remain rackless and didn't find out until it was too late? For Pioneer-lovers, at least, I can bring news of a light at the end of the tunnel. Shriro (UK), the importers of Pioneer equipment, have told me that Pioneer racks will be available as separate units in about five months.

## Keeping the Homeleds Burning

Battery powered equipment is a nuisance. More than that is a guilty-of-incestuous behaviour nuisance. Mainly because I keep leaving it on and flattening the batteries - and on my headlamp that means two days without moving-coil type music - leading to withdrawal symptoms such as shortened tempers and kicked cats

What is needed is a machine which knows when you're using the box and when you're not - and makes an appropriate fuss if you leave it on. After several periods of non-musical evenings I finally overcame my inertia and DID SOMETHING ABOUT IT. Some moons ago ETI ran a project called the Watchdog which was designed to prevent you leaving on teles, hi-fis and other (mains driven) equipment. This was also born of my desire to be lazy electronically.

penes sey ejey дenoy 'ı!

 de-power the equipment in question if need
less applicable to battery powered devices.
$\qquad$
 consideration. DI is protection against back


 sient audio, such as switctrof thamps e
re-triggering the Watchdog and keeping from closing down.

SKI feeds the input audio of around 50 mv minimum to the buffer amp formed by 1 Cla and associated components. The gain of this and may be changed to vary the unit's sensitivity. ICID is a second order Bessel low pass
fier with break frequency set around 800 Hz (when loaded). This is to cut the response to high frequencies such that continuous white



 square-wave,


With a little modification Watchdog can be amended to sound an audible alarm. The circuit is given below, and note that, in order that the box is 'fail-safe', the switching relay is under power as long as the equipment is operating. This means that rather than replacing the relay with an audible 'bleeper' or the like 'twould be better to wire up one set of contacts to power said device once the relay tripped options open - except on cold days. ETI

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| RESISTORS ( $1 / 4 W \mathrm{~W}$ (12) |  |  |  | 2N5777 | 55p | LM308N | 600p | 4049 | 35p | 7460 | 13p | 74195 | ${ }^{57 p}$ | BF196 | 12p | ZTX504 | 28p |
| PRESETS (.15W Horizontal) |  |  |  | OCP71 | 65p | LM318N | 200p | 4050 | 35p | 7470 | 28p | 74196 74197 | 100p | BF197 | 12 p | 2N696 | 35p |
| 100 ohms to 2 Mohms ${ }^{\text {POTENTIOMETERS (Carbon) }}$ |  |  |  | ORP 12 | 60p | LM339N | 60p | 4066 4069 | $38 p$ $\mathbf{1 8 p}$ | 7472 7473 | 21p | 74197 74198 | $100 p$ 1350 | ${ }^{\text {BF244B }}$ | 35p | 2N697 | 25p |
|  |  |  |  | MAN 72 | 110p | LM348N | 90p | 4070 | 18 p | 7474 | 240 | 74199 | 120 p | 8 BFR 39 | 32p | 2N698 | 35p |
| Log and Linear |  |  |  | MAN74 | 110p | LM377N | 175p | 4071 |  | 7475 | 34p |  |  | BFR79 | 32p | 2N706 | 14p |
| 4.7 Kohm to 2.2 Mohms VERO BOARDS (0.1" Copper) |  |  |  | 125" |  | LM380N | 90p | 4072 | 18 p | 7476 | 30 p | transist | ORS | BFX29 | 25p | 2N914 | $20 p$ |
|  |  |  |  | Ds |  | LM381N | 140p | 4073 | 18 p | 7480 | 32 p | AC126/7 | 23p | BFX84 | 25p | ${ }^{2} \mathrm{~N} \mathrm{~N}^{2} 18181$ | 35 |
| $2.5 \times 5$ 52p |  |  |  | Red | 10p | LM382N | 130p | 4081 | 18p | 7485 | 80p | AC128 | 23p | BFX87 | $25 p$ | 2N1131 | 200 |
| $375 \times 5$ZENER DIODES ( 400 mW ) 62 |  |  |  | Green | 4p | LM1310N | 150p | 4082 | 18p | 7486 | 25p | AC128/176 |  | ${ }_{8} \mathrm{~F} \times 88$ | 25p | 2N1302 | 35p |
|  |  |  |  | Yellow | 149 | LM3900N | 55p | 4086 | $80 p$ | 7490 | 36 p | Mt.pr. | 42p | 8FY50 | 22p | 2N1303 | ${ }_{50 \mathrm{p}}^{40 \mathrm{p}}$ |
| TRANSFORMERS (240V) 8p |  |  |  | ${ }^{12125^{\prime \prime}}$ clip | 3p | LM3909N | 70p | 4510 | 73p | 7491 | 55p | AC176 | 23p | BFY51 | 22p | 2N1613 |  |
|  |  |  |  | ${ }^{2 \prime}$ clip | 4 p | MC1496P | 85p | 4511 | 99p | 7492 | 40p | AC187/8 | 23p | BFY52 | 22p | 2N2222A | 25p |
| 9.0 .9 V .100 mA0.15 V .015 V 200 mA |  |  | 130p | DIODES |  | NE531 | 140p | 4516 | 95p | 7493 | 35p | AD149 | 70p | 8RY39 | 60p | 2N2369 |  |
|  |  |  | 150p | 8 YY 127 | 10p | NE555 | 25p | 4518 | 72p | 7494 | 57p | AD161/2 | 40p | BSX20 | 22p | 2N2484 | 30 p |
|  |  |  |  | OA47 | 8 pp | NE556 | 60p | 4520 | $94 p$ | 7495 | 46p | AF124 | 45p | 8U205 | 1500 | ${ }^{2} \mathrm{~N} 2484$ | 55p |
|  |  |  | 2p | 0491 | 8 p | NE566 | 140p | 4528 | 99p | 7496 | 56p | AF139 | 40p | BU208 | 210p | 2N2904 | 23p |
| POLYSTYRENE CAP (50V) |  |  |  | OA200 | 6 p | tba64 1A | $200 p$ |  |  | 7497 | 200p | AF239 | 47 p | MJ2955 | 110p | 2N2905 | 23p |
| 100F to 1.000 pF |  |  | 5 p | OA202 | 9 p | tBA641B | 200p | TTL |  | 74100 | 90p | BC107 | 10p | MJE340 | 70p | 2N2906 | 200 |
|  |  |  |  | ${ }^{1} \mathrm{~N} 916$ | 5p | TBA800 | 75p | 7400 | 13p | 74105 | 43p | BC108/9 | ${ }^{10} \mathrm{p}$ | MJE2955 | 110p | 2N2907 | $20 p$ |
| POLYESTER CAP ( $\mathbf{1 0 0 V}$ ) <br> 1 nF to 100 nF |  |  | 5 p | 1 N4148 | 4p | TBA810S | 110p | 7401 | 13p | 74107 | 26p | BC147 | 10p | MJE3055 | $85 p$ | 2N2926G | 11p |
| -15, 22, 33, $39 \mu \mathrm{~F}$ |  |  | ${ }_{6 p}$ | 1 N 4001 | 4 p | ZN414 | 100p | 7402 | 13 p | 74109 | 44p | BC148 | $10 p$ | MPF102 | 40p | ${ }_{2 N}{ }^{\text {N2926G }}$ | 11 p |
| 47, $68 \mu \mathrm{~F}$ |  |  | 10p | 1 N4002 | 4 p | 2N1034 | 200p | 7403 | 13p | 74110 | 40p | BC149 | 10p | MPF 103 | 40p | 2N3054 | 50p |
| $1 \mu \mathrm{~F}$ | 12p | $2.2 \mu \mathrm{~F}$ | 20p | 1 1N4003 | 5 p |  |  | 7404 | 13p | 74118 | ${ }^{90} \mathrm{p}$ | BC157 | 12p | MPF104 | 40p | 2N3055 | 50p |
| $3.3 \mu \mathrm{~F}$ | 26p | $4.7 \mu \mathrm{~F}$ | 28p | 1 N 4004 | ${ }^{6 p}$ | CMOS |  | 7405 | 13p | 74121 | 26p | BC158/9 | 12p | MPF 105 | 40p | 2 N 3442 | 140 p |
|  |  |  |  | 1 N 4005 | ${ }^{6 p}$ | 4000 | ${ }_{18 \mathrm{p}}^{18}$ | 7406 | 24p | 74122 | 40p | ${ }^{\text {BC167 }}$ | 14p | MPF 106 | 50p | 2N3702 to |  |
| 1/2510 47/25 ${ }^{\text {che }}$ ( $\mu \mathrm{F} / \mathrm{V}$ ) |  |  | 6 p | 1 N 4006 | 8 p | 4001 | 18 p | 7401 | 24p | 74123 | 45p | BC169C | 13p | MPSA06 | 26p | 2N3711 ${ }^{\circ}$ |  |
|  |  |  | 8 p | 1 N 4007 | 8p | 4002 | 18p | 7408 | 13p | 74125 | 37p | ${ }^{\mathrm{BC}} 173$ | 8 p | MPSA56 | 26p | 2N3772 | $150 p$ |
| 68/50. $100 / 35$$150 / 25.200 / 12$ |  |  | 9p | 1 N 5400 | 13p | 4006 | 70p | 7409 | 13p | 74126 | 37 p | BC177/8 | 18p | MPSU06 | $61 p$ | 2N3773 | 250 p |
|  |  |  | 10 p | 1 N5401 | $14 p$ | 4007 | 18p | 7410 | 13p | 74132 | 48p | ${ }^{\mathrm{BC}} 179$ | 20p | OC35 | 92 p | 2N3819 | 22p |
| $\begin{aligned} & 220 / 25.250 / 12 \\ & 470 / 25.500 / 30 \end{aligned}$ |  |  | 13p | 1N5402 | 15p $16 p$ | 4008 4009 |  | 7411 7412 | 18p | 74141 74145 | 60p 550 | ${ }_{\text {BC184 }}{ }^{\mathrm{BC} 182 / 3}$ | 12p | TIP29 | $40 p$ | 2N3820 | 40p |
|  |  |  | 14 p 22p | 1N5404 | 16p | 4010 | $34 p$ $42 p$ | 7412 7413 | 15p 27p | 74145 74150 | 55p 78 p | BC184 BC209 | 12p | ${ }_{\text {TIP30 }}$ | 48 p | 2N3823 | 70p |
| $\begin{aligned} & 1000 / 25 \\ & 1500 / 25 \end{aligned}$ |  |  | ${ }_{26 p}$ | DIL SOCK |  | 4011 | 18p | 7414 | 31p | 74151 | 48p | BC212/3 | 12 p | TIP308 | 48 p | 2N3866 | 90 p |
| 2200163 |  |  | 20p ${ }^{\text {' }}$ | ${ }^{8} 8 \mathrm{pin}$ | $11 p$ | 4012 | 18p | 7416 | 25p | 74153 | 43p | BC214 | 12 p | TIP31 | 40 p | 2N3903 | 10p |
|  |  |  |  | 16 pin | 149 | 4013 | 40p | 7417 | 25p | 74154 | 90p | BC214L | 14p | TIP32 | 40p | 2 N 3905 | $10 p$ $10 p$ |
| BRIDGE  <br> RECTIFIERS VOLTAGE <br> REGULATORS  |  |  |  | 18 pin | 18p |  | ${ }_{70}$ | 7420 | 14p | 74155 | 56p | BC261B | 14p | IIP33 | 60p | 2N3906 | 10p |
|  |  |  |  | 22 pin | 22p | 4015 | 70p $40 p$ | 7421 | 17 p | 74156 | 46p | BC461 | 40p | TIP33C | 80 p | 2N4037 | 45p |
| 0.75/200V 25p |  | 32 OH 05 | 40p | 24 pin | $24 p$ | 4017 | 47p | 7422 7427 | 17p | 74157 74160 | 43p | BC477 BC478/9 | ${ }_{\text {27p }}^{\text {27p }}$ | TIP34A |  | 2 N 4058 | $14 p$ |
| $0.75 / 600 \mathrm{~V}$ 30p |  | $32 \mathrm{OH}-24$ | 40p | 28 pin | 28p | 4018 | 68p | 7427 7428 | 20p | 74160 74161 | 64 p 64 p | BC478/9 | 27p | ${ }_{\text {TIP358 }}$ | 240p | 2N4059 | 14p |
| 1A/50V 22p |  | 7805 | 70p | 40 pin | 40p | 4019 | 45p | 7428 7430 | ${ }^{25 p}$ | 74161 | 64p | BC547/8 | 14p | TIP36B | 280p | 2N4060 | 14p |
| $1 \mathrm{~A} / 100 \mathrm{~V}$ | 27p | 7812 | 70p | linear |  | 4020 | ${ }_{900}$ | 7430 | $14 p$ | 74162 | 64p | BC549 | 14 p | TIP41A | ${ }^{60 p}$ | 2N4061 | 14p |
| $1 \mathrm{~A} / 200 \mathrm{~V}$ 32p |  | . 7815 | 70p | CIRCUITS |  | 4021 | 85p | 7433 | 24p | 74164 | 64 p 78 p | ${ }_{\text {BC559 }}$ | 15 | TiP42A | 60 p | 2N5457 | 40p |
| $\begin{array}{ll}1 \mathrm{~A} / 400 \mathrm{~V} & \mathbf{3 4 p} \\ 2 \mathrm{p} / 50 \mathrm{p} & 40 \mathrm{p}\end{array}$ |  | 7818 | 70p | 709 | 40p | 4022 | 82p | 7433 7437 |  | 74164 74165 | $78 p$ $78 p$ | ${ }_{\text {BC5 }} \mathrm{BC} 59$ | $15 p$ $18 p$ | TIP2955 |  | 2N5458 | 40p |
|  |  | 7824 | 70p | 710-14 | 33p | 4023 | 18 p | 7438 | 20 p | 74166 |  | ${ }_{\text {BCY71 }}$ | 18 p | 1P305 | 55 p | 2N5459 | 35p |
| $\begin{array}{ll}\text { 2A/ } 100 \mathrm{~V} & \text { 42p } \\ 2 \mathrm{~A} / 200 \mathrm{~V} & 48 \mathrm{p}\end{array}$ |  | 7905 | 90 p | 741.8 | 22p | 4024 | 58p | 7440 | 20p | 74173 | 87p | ${ }_{\text {BCY7 }}$ | 18p | 218107 | 13p | 2N6027 | 40p |
|  |  | 7912 | $90 p$ | 747-14 | 48p | 4025 | 18p | 7441 |  |  |  |  |  |  | 13p |  |  |
| 2A/400V | 55p | 7915 | 90p | 748-8 |  | 4027 | 35p |  | 56p | 74175 | 75p | BD131/2 | 58p | 21x109 | 13 p |  |  |
|  |  | 7918 | 90p | CA3018 | 70p. | 4028 | 60p | 7442 <br> 7443 | 51p | 74176 7417 | 64p | ${ }_{\text {BDD }}$ | 42p | 218300 | 16p |  |  |
|  |  | 7924 | 90p | CA3028A | 85p | 4029 | 75p | 7444 | 80p 80 p | 74176 74177 | 64 P $\mathbf{6 5 p}$ | BDD BD137/6 | 42p | $21 \times 301$ LTX | 18p |  |  |

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HOW IT WORKS


Lay down the ETIPRINT and rub over with a soft pencil until the pattern is transferred to the board. Peel off the backing sheet carefully making sure that the resist has transferred. If you've been a bit careless there's even a 'repair kit' on the sheet to correct any breaks!

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# Henry Budgett has news of the long-awaited Nascom 2, a pretty potent Petsoft program party and an expensive Communicator. 

IF micros are the food of life eat on would be a fair description of this month's ramblings through the world of small systems. The spirit of Christmas also seemed to take quite a bashing as well! The old adage about it being human to err has struck again. Yours truly has made a box-up. Having rambled extensively on the subject of standards in the November offering, Mr Henry Law has raised the valid point about the usage of the word Baud I tend to use both Baud and bits per second to mean the same thing and he points out quite rightly that they don't. The term "bits per second" means exactly what it says, but the term Baud is defined as being "a unit of signalling speed equal to the number of code elements transmitted per second." Some transmission systems such as phase modulation actually carry more than one bit in each signal element so in Mr Law's case his 9600 bps transmission runs at 2400 Baud. In future l'll stick to bps unless the actual Baud rate is defined, honest.

## The Long Wait (Is Over)

The first gastronomic offering of the month was digested after the delivery of our Nascom 2. Not bad really, we have been waiting since June along with the rest of you for the machine. The hardware is certainly impressive. There are options galore for both memory and 1/0 and the packing density of chips is almost unbelievable. The lack of 4118 memory chips has caused Nascom to offer the 16 K RAM board with the " 2 " but if you wish to use it in a dedicated system the missing 4118 s can be replaced with 2708 EPROMs holding your own software. The BASIC is the 'standard' Microsoft version, but it has one or two added extras to make periperheal handling easier.

The other interesting feature is that the BASIC is blown in a 64 K ROM chip, one of the first uses for this device, and is actually removable so you can use the memory space for something else. Both the new NAS-SYS monitor and the BASIC can be used on the " 1 " and I feel that the NAS-SYS should become the standard operating system for the " 1 " rather than the T2 or T4.

The instruction manual is a loose leaf binder making it easy to alter or insert new information. There are one or two poor areas, but Nascom have promised to change these if demand is sufficient. The price may seem high but you are getting a very high quality product that makes certain other "single board computers" look rather like toys.

A very strong whisper on the Nascom front is that the price of the " 1 " may well be coming down, look out for bargains. The lunch was very good but it's the first time l've had to pay for doing someone a review of their own product!


Would you buy 50,000 programs from this man?

## The Second Belch

The second gastronomic wonder was given by Petsoft to celebrate, among other things, the sale of the 50.000 th program. They have even had it gold plated and presented to the author. I rather suspect that the dreaded LOAD ERROR message may come up if it is ever used. Gathered at the lunch were many of the editors and publishers of the Computing Press and after the dinner discussions ranged far and wide on the subject of machines and languages. As usual in these meetings nobody managed to agree on anything apart from the excellence of the wine and the quality of the food!

The fact that Petsoft has become a part of the giant ACT doesn't seem to have detracted from the quality of their product. If anything, it has dramatically improved. The achievement of selling 50,000 programs has only left them with one slight problem, do they go for platinum plating on the 100,000 th?

## Fresh Air

We have had a small box called the Communicator, in the office for review. The device hangs on the parallel user port of the PET and provides eight user accessible input or output lines. The voltage of the power supply that you connect determines the output swing. The box derives five volts internally for the input lines. The circuit is based around two Darlington transistor arrays, each containing seven transistors that are connected to provide two channels with a drive capability of 500 mA and six channels that can drive 1 A . The box can be directly called from BASIC
using either POKE for output or PEEK for input. The manual is not too specific, but interrogation of the PET manual, the 'one with the blue writing on the cover, reveals that there are a few extra things that you can do. Unfortunately the unit has no handshaking capability at all so most of your program will be taken up with loops to interrogate the device.

The other contention that I have with the device is that for a grand total of $£ 92.85$ including VAT and p\&p you don't really get value for money. The Communicator is available from Mektronic Consultants, Linden House, 116 Rectory Lane, Prestwich, Manchester M 25 5DB


The Communicator performs well, but does it represent value for money?

## Exhibitionitis Strikes Again

There was one ray of sunshine in the gloom of computer shows this month, no it wasn't Compec. The Professional Viewdata Exhibition held at the West Centre Hotel last month was the one that rates high on my list for a visit next time. A grand total of 28 exhibitors, all doing their utmost to be helpful, plenty of space to move around, comfortable seats for tired journalists - in fact everything you could wish for. Among the notable were Tecnalogics who are doing very well with their TECS system, Logic Box with their screen treatments - amber is the IN colour - and Video Electronics with their intriguing real time graphics display system.

Make a note in your diaries to go to this one next year, it's well worth it. The same cannot be said of Compec which was incredibly hot, crowded and boring. A companion of mine, who was there to check out some hardware, reckoned the pub was less crowded at lunchtime, so we all went there instead. About the only points of interest were the new Sharp machine, but no-one seemed to be very impressed, and the Texas machine which is expensive. One day someone will organise a true home computer show with no big business computers, lots of room to walk around and a friendly atmosphere in which one can actually talk to the people you've gone to see. Surely it's not too much to ask.

ETI

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ELECTRONICS TODAY INTERNATIONAL - FEBRUARY 1980

# ambit 

INTERNATIONAL

# IT'S HAPPENIED RCAII! THE PART THREE CATRLOCUE IS PUBLISHED \& We have moved to bicier premises. 

Yes. it, here at last - the all new Part Three Catalogue. Fun for all the family, and the usual update on all that is new, worthwhile and exciting in the world of Radio and Communcations. A big section on freguency svethesis techmiques covering broadeast tuners. to communication guality transmitter systems. Hore new products than ever- RADIO CONTROL parts. crystal filters. ceramic filters for 455 hH , and the new range of TOKO (FSH low temperature coefficient types for 10.7 MH . Details on new radio ICs . including the new HAll225. the ( A. 3189 E E lookalike with 84 AB signal to noise, and adjustable muting threshold. Radio comtrol ICs - and an updated verion of the RCM\& 8 chamel FM receiver now with an Ambit designed screened front end. with 27 MH , ceramic bandpass filter. LCD panel cloch timer modules. the neatest amd best LCD pancl DVM vet conly 19.45 each + VAT), the new 5 decade resolution DFM3 for LW:HF VHF with LCD readout. The DF M6 with fluorescent display 1010 kH , resolution on VHF. 1 hHz on SW . A lkH , HF synthe siser with five ICs. the list is endless. Get your copy of the catalogue now Post publication price is 60 p tine PP ete). The previous two sections are also reguired for a complete picture: Parts 1 \& 2 fl the pair. All 3 © 1.50 And don't miss our spot the gibbon contest, together with a guiz to see if you can spot the differences between a beolithic cave drawng and a circuit diagram of one of our competitor's tuners. * Yes. we still haven't learmt how to spell.)



Not illustrated here - but also now available is the DFM6. This is a vacuum fluorescent display version of our immensely popular DFM3 (LCD). Resolution is 100 Hz to $3.9999 \mathrm{MHz}, 1 \mathrm{kHz}$ to 39.999 MHz , and 10 kHz to $200.00 \mathrm{MHz}^{+}$; all standard IF offsets (inc. 10.7 MHz on shortwave)

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OUR LATEST MOVING EXPERIENCE :: At last, we have moved to the address below. There is car parking for customers approaching via North Service Road (an extension of North Road Avenue, entrance opposite the Brentwood Fire Station.) Pedestrian access from the High Street (alongside 117 High Street). The new building is six times bigger than our Gresham Road offices, and we will be installing a much expanded sales counter in the fullness of time. NEW TELEPHONE NUMBER (0277) 230909, TELEX NUMBER (as before) 995194 AMBIT G. See you there


## Traffic Light Controller

Michael Miller

This circuit is relatively simple and gives a realistic timing sequence. IC1 sets the timing clock pulse and can be adjusted by RV 1 . IC2 is a decade counter, whose output pulses are mainly fed through diode buffers any small cheap diodes will do), to IC3, a quad OR gate, which sorts the consecutive decade pulses into three groups, monitored by the three coloured LEDs

To couple this circuit to a similar one, for the other intersection of the crossroads, the pulse from pin 1 of IC2 should be taken to pin 15 of the IC2 of the other circuit. This second circuit should have pin 15 biased to OV via a 100 k resistor. When the first circuit is showing red, the second circuit will be showing green.

Tech-Tips is an ideas forum and is not aimed at the beginner. We regret we cannot answer queries on these items.

ETI is prepared to consider circuits or ideas submitted by readers for this page. All items used will be paid for. Drawings should be as clear as possible and the text should preferably be typed. Circuits must not be subject to copyright. Items for consideration should be sent to ETI TECH-TIPS, Electronics Today International, 145 Charing Cross Road, London to ETI TECH
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## Silent Sentry

B. J. Lowery

The 'Silent Sentry' is a form of intrusion alarm. It will indicate the breaking of a light beam by means of lighting a Light Emitting Diode, which will remain lit until the RESET switch is activated. Q1 may be any suitable NPN photo-transistor available, eg 276-130 (Tandy) and BPX 25 (Maplin).

When light from the light source is falling on Q1, Q2 is turned on, causing both Q3 and Q4 to be turned off, therefore LED1 will not light. As soon as the light beam is broken Q2 is turned off, allowing Q4 to be turned
on via RV1. LED 1 lights, Q 3 is turned on via R2, SW1, LED 1 and Q4. Q3 forms a latching circuit for Q4 and this will keep LED 1 alight until the RESET switch is operated. The GAIN control is a sub-miniature preset and should be adjusted to give LED 1 a positive on / off action.


(RS 196-369)

## Battery Charger Controller <br> D. Wedlake

The battery charger circuit illustrated was designed to be incorporated in any conventional battery charger rated up to 10 amps , where the output is full-wave rectified and unsmoothed. It is fully protected as it cannot be damaged by short circuit or reverse battery connection. Furthermore, charging ceases when the battery voltage reaches a pre-set voltage (normally 13 V 8 for a fully charged battery).

The design is based on the Programmable Unijunction Transistor (PUT) oscillator which senses the
battery voltage to determine when charging should cease. The battery being charged provides the power for the oscillator which, in turn, triggers the thyristor via the pulse transformer T1. As the anode of the PUT is clamped to 5 V 6 by the Zener Diode, ZD1, it follows that the circuit will not oscillate if the potential at the slider of RV1 is correspondingly higher. Therefore, RV1 controls the cut-off point which should be set to 13 V8. This is best set under actual operating conditions and the charging current will gradually reduce as this voltage is approached.

The charger is fully protected as the circuit cannot oscillate under short circuit conditions or reverse battery connections. However, as the power for the oscillator is derived
from the battery, the circuit will naturally not be self starting if the battery is completely flat or charged to less than about 7 volts. This problem could be overcome by providing a push-button shorting switch across the thyristor to initiate charging. In a short while the battery voltage should have risen sufficiently to maintain normal operation. However, one should bear in mind that the charger will not be protected when the start push button is pressed, so if included, one should provide a fuse as additional protection.

If used at full load current, the thyristor should be mounted on a suitable heat sink having a thermal dissipation of $4 \mathrm{C} / \mathrm{W}$;eg RS 401497).

## 8K ON BOARD MEMORYI

$5 K$ RAM. 3 K ROM or $4 K$ RAM, 4 K ROM (link selectable). Kit supplied with 3 K RAM, 3 K ROM System expandable for up to 32 K memory.

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Cabinet size $19.0^{\prime \prime} \times 15.7^{\prime \prime} \times 3.3^{\prime \prime}$. Television by courtesy of Rumblelows Ltd., price $£ 58.62$

2 MICROPROCESSORS
280 the powerful CPU with 158 instruction, including all 78 of the 8080 , controls the MM57109 number cruncher. Functions include,,+- 1 , squares, roots, logs ${ }^{\text {g }}$ xponentiags, trig functions, inverses etc. Range $10^{-95}$ to $9 \times 19^{99}$ to 8 figures plus 2 exponent digits.

## EFFICIENT OPERATION

Why waste valuable memory on sub routines for numeric processing? The number cruncher handles everything internally!

RESIDENT BASIC
with extended mathematical capability. Only 2 K memory used but more powerful than most 8K Basics!

1K MONITOR
resident in EPROM.
SINGLE BOARD DESIGN
Even keyboards and power supply circuitry on the superb quality double sided plated through-hole PCB.

## PONERRAM

## PSI Comp 80. 280 Based powerful scientific computer Design as published in Wireless World, April-September, 1979.

The kit for this outstandingly practical design by John Adams published in a series of articles in Wireless World really is complete Included in the PSI COMP 80 scientific computer kit is a professionally finished cabinet, fibre-glass double sided, plated-through-hole printed circuit board. 2 keyboards PCB mounted for ease of construction, IC sockets, high reliability metal oxide resistors, power supply using custom designed toroidal transformer. 2 K Basic and 1 K monitor in EPROMS and, of course, wire, nuts, bolts, etc.

## PSI COMP 80 Memory Expansion Sys-

 tem| Expansion up to 32 K all inside the computer's own cabinet! |  |
| :---: | :---: |
| By carefully tho own power sup 38 K RAM or cabinet Conne socket is made | gineering a mother board with buffers and its by the computer's transformer) enables up to ards to be fitted neatly inside the computer mother board from the main board expansion cable. |
| Mother board: | Fibre glass double sided plated through hole P.C.B. $8.7^{\prime \prime} \times 3.0^{\prime \prime}$ set of all components including alt brackets, fixing parts and ribbon cable with socket to connect to expansion plug <br> $£ 39.90$ |
| 3K Static RAM board | Fibre glass double sided plated through hole P.C.B $5.6^{\prime \prime} \times 4.8^{\prime \prime}$ <br> £12.50 <br> Set of components including IC sockets, plug and socket but excluding RAMs <br> £11.20 <br> 2114 L RAM ( 16 required) $£ 5.00$ <br> Complete set of board, components, 16 RAMS <br> $£ 89.50$ |
| 8K ROM board | Fibre glass double sided plated through hole PC.B. $5.6^{\prime \prime} \times 4.8^{\prime \prime}$ <br> $£ 12.40$ <br> Set of components including IC sockets, plug and socket but excluding ROMs . . £10.70 <br> 2708 ROM ( 8 required) <br> $£ 8.00$ <br> Complete set of board, components, 8 ROMs £78.50 |

Floppy Disk, PROM programmer and printer interface coming shortly!

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## Stereo VCA

J. Macaulay

The circuit shown is of a stereo VCA whose gain can be varied over a 90 dB range by the application of a contro! voltage between $0-15 \mathrm{~V}$.

Maximum gain is limited to 20 dB and occurs when the control voltage is OV . Minimum gain occurs with the application of +15 V at the control input.

The circuit works as follows. IC1 / 2 are 741 op amps operated in the virtual earth mode with R1, R5 determining the input impedance at 1 M , regardless of gain. The feedback loop from the output of the IC's are completed by the resistors R4, R6. A pair of MOSFETs, internal to IC3, are connected in parallel with these resistors and the control voltage is applied to their gates, pins 3 and 10 .

When zero volts are applied to the gates the resistance across the feedback loop is some $10^{9}$ ohms in all with R4-6. In consequence these latter components determine the gain of the stage. When the control voltage is increased in a positive direction the impedance across the


FETs decreases and the gain of the amplifier decreases in sympathy. Once the voltage is increased to 15 V the impedance across the FETs lowers to roughly 300R.

The frequency response of the amplifier extends from approximately $5 \mathrm{~Hz}-100 \mathrm{kHz}$ at the
-3 dB points whilst the distortion at maximum gain is about $0.1 \%$ at 1 kHz . If the feedback resistors are close tolerance types, $2 \%$, the gain will be found to be within $\pm 1 \mathrm{~dB}$ between channels due to the closely matched characteristics of the FETs within IC3.

## LED Chaser

P. Davidsu

This game is a test of skill and patience. The aim is to align a LED chaser (under your control), with another preset chaser.

A matching pair of outputs are fed
to an AND gate (IC4a, IC3e). This gate feeds the NAND gate 1 C 4 b , its other output taken from the monostable formed around IC3d. This has a duration of about 6 seconds to ensure the display is fully counted.

When you think you've matched
the displays up PB1 is pressed. IC3d output goes high and if, while this is high, the two matching outputs both go high, the monostable formed around IC3f is triggered. This enables the astable formed by IC4c, d, to signal success. The unit can also be automatically reset by feeding IC3f's output to the clock inhibit pin of IC2 or 1).


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## PCB <br> FOIL PATTERNS

A small departure this
 month. We are giving the front panel layouts for the Project 80 modules, just to see whether or not it is really of use to our readers. Let's have your opinions.



[^0]:    Electronics Ioday International is normally published on the first Friday of the month prior to the cover date

[^1]:    Below: the complete Casio package. The two phantom keyboards to the right of the FA 1 are the overlays for programmable and music playing usage. Note that the FA1 is an optional extra and you don't get it free with the 502P (shame shame . . .)

[^2]:    Personal Shoppers EDGWARE ROAD LDNDON W2 Tel: 01 -723 8432. 9.30am-5.30pm. Closed all day Thursday ACTON: Mail Order only. No callers goods despatched tomainland and w. IRELano only

[^3]:    The items mentioned here are those planned but circumstances may affect the actual contents

[^4]:    PO BOX 23, 34 SEAFIELD ROAD, COPNOR, PORTSMOUTH, HANTS. PO3 50J
    CALCULATOR CHIP NOHEC 4204 functon and K WIth data 850 . MME314 Digital chip with dath $\mathbf{E 1 . 9 9}$ each MM5316 Digital aldrm clock chip with dara $\mathbf{E 2 . 4 9}$ each. DIGITAL ALARM CLOCK MODULE With O 8 ted dISplay With data EE.50 each \& DIGIT O.f"LED DISPLAY Common cathode With data E3.95 each LED WRISTWATCH I.C. Masten MK
     The MK 5030 and 015501 whe packagel in iegless tlatpack format and require some farly fin Solitering 20 KEY KEYBOARDS calrulator style keyboards 2 for $£ 1.00$. MM 2102 MEMORIES TIMETER CHIP MMS 330 IC O.1"LED DISPLAY Conn on Cathorie 99p. IMS5S TIMER IC. With data each. 8 DIGIT booktrt 25p mach SLIDER SWITCHES 2 pole thange.over 16 p मach PUSH BUTTON SWITCHES sprilg inated tinmientary with ino contact 15p each POLARIZING
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