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## 回回回 THE

## THE FABULOUSLY SUCCESSFUL VISCOUNT III AUDIO £52complete <br> $14+14$ watts r．m．s． 40 Hz to 40 kHz 3 dB ．Total distortion at 10 watts

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$0 / 60 \mu \mathrm{~A} / 6 / 60 / 600 \mathrm{~mA}$. $0 / 60 \mu \mathrm{~A} / 6 / 60 / 600 \mathrm{~mA} . \quad 0 / 6 \mathrm{~K} /$
$600 \mathrm{~K} / 6 \mathrm{Meg} . / 60 \mathrm{Meg}$.


MODEL TE-200 20,000 O.P.S. Mirror scale overload $0 / 10 / 50 / 250 / 1,000 \mathrm{~V}_{\text {a.c. } 0 / 501 \mathrm{LA}}$ $250 \mathrm{nAA} .0 / 60 \mathrm{~K} / 6 \mathrm{meg} \Omega .-20 \mathrm{t}$
+62 dB . 29.95 P. $\&$ P. 10 p.


HODEL 50030,000 O.P.V with overload protection mirror scale $0 / 0 \cdot 5 / 25 / 10 / 25 /$ $100 / 250 / 500 / 1.000 \mathrm{~N}$
$0 / 2.5 / 10 / 25 / 100 / 250 / 500 /$ $1,000 \mathrm{~V}$ a.c. $0 / 50 \mu \mathrm{~A} / 5150 /$
 88.87 . Port paid.
 10 amp. a.c.


$0 / 5 \mathrm{~K} / 50 \mathrm{~K} / 50 \mathrm{~K} / 5 \mathrm{MEG} / 50 \mathrm{MEG}$.
$-20+62 \mathrm{~dB} .215$, P. \& P. $2 \overline{\mathrm{p}}$.



## Selected TEST EQUIPMENT


$\begin{array}{lr}\text { MODEL } & \text { PL438 } \\ 20 \mathrm{k} \Omega / \mathrm{V} & \text { a.c. } \\ 8 \mathrm{k} \Omega / 5\end{array}$
a.c. Mirror scale.
$0.6 / 3 / 12 / 30 / 120 / 600 \mathrm{y}$
d.c. $3 / 30 / 120 / 600 \mathrm{~V}$
A.c. $\quad . \quad .0 / 600 \mu \mathrm{~A} / 00 /$
600 mA.
$10 / 100 \mathrm{~K} /$
-00 to $+45 \cdot 113$. 86.97 , P. \& P. 12 p .


## MODEL

Taut band
suspension.
Orerloar pr tection. Polarity re-
vering switch.
30.0000 P
$0 / 0.5 / 2 \cdot \bar{j}$ i 15

$50 / 250 / 500 / 1,000 / 2,500$ 大ु d.c. $0 / 15 / 50 / 150 /$ $500 / 1,000 \mathrm{~V}$ a.c. $0 / 50 \mu \mathrm{~A} / 5 / 50 / 1 \mathrm{so/500mA/}$
$\overline{\mathrm{JA}}$ d.c. $0 / 3 \mathrm{~K} / 300 \mathrm{~K} / 3 \mathrm{meg}$. $88-85$. Post 20p.


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$0 \cdot 3 / 0 \cdot 6 / 1 \cdot 2 / 1 \cdot 5 / 3 / 6 / 12 / 30 / 60 /$ $120 / 300 / 600 / 1,200 \mathrm{~V}$ t.
$1 \cdot \mathrm{O} / 3 / 6 / 12 / 30 / 60 / 1$ $1 \cdot 5 / 3 / 6 / 12 / 30 / 60 / 150 / 300 /$ $600 / 1,200 \mathrm{~V}$ z.c.
$15 / 30 \mu \mathrm{~A} / 3 / 6 / 30 / 60 / 1 \nu 0 / 300 \mathrm{~mA}$
$6 / 12$ amp. 3. Mes/ P. \& Meg oh 20 p


MODEL C-7080 EN Giant 6 in mirror reale.
20,000 O.P.V. $0 / 0 \cdot 25 / 1 / 2 \cdot 5 / 10$ $1,000 / 5,000 \mathrm{~V}$ d.c. $0 / 2 \cdot \mathrm{~J} / 10$ $150 / 2 \overline{0} 0 / 1,000 / \bar{a}, 000 \mathrm{~V}$ a.c. $0 / 50 \mu \mathrm{~A} / 1 / 10 / 100 / 500 \mathrm{~mA}$ $10 \mathrm{amp} .1 . c .0 / 2 \mathrm{~K} / 200 \mathrm{~K} /$ 20 meg.
-20 to


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 Extremely sturdy instr for general667 O.P.V.
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$0 / 300 \mathrm{~L}$
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| :--- | :---: | :--- |
| Amps | Volts |  |
| 10 | 60 | PM7A1 |
| when | 125 | PM7A2 |
| mounted | 250 | PM7A4 |
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## THE HY41



The HY41 supersedes the popular HY40 introduced by ILP last year. This highly improved module achieves true High Fidelity with a dramatic reduction in distortion Itypically $0.05 \%$ at 1 KHz into 8 ohms! and is electronically and mechanically compatible with the HY40

With this important improvement the HY 41 retains all of the quality characteristics found in the earlier version and P.C. board, Resistor, Capacitors, Hardware Mountings and comprehensive manual are included in the basic kit. No further components are required to construct a complete power amplifier of extremely high performance sufficiently versatile to provide power not merely for Hi-Fi but also for public address systems and industry.

The free manual gives a full circuit diagram of the HY41 and its various applications including complete stereo amplifier.

Like its predecessor the HY41 is based on conventional and proven circuit techniques developed over recent years.
OUTPUT POWER: British Rating 40 WATTS PEAK, 20 watts
R.M.S. continuous.

LOAD IMPEDANCE: $4-16 \mathrm{ohms}$
INPUT IMPEDANCE: 30 K ohms at 1 KHz
VOLTAGE GAIN: 30 db at 1 KHz
TOTAL HARMONIC DISTORTION: less than $0.15 \%$ (iypical $0.05 \%$ )
at 1 KHz
FREQUENCY RESPONSE: $5 \mathrm{~Hz}-50 \mathrm{KHz}+1 \mathrm{db}$.
SUPPLY VOLTAGE: + $22.5 v o l t s$ D.C
SUPPLY CURRENT: 0.8 amps maximum

PHICE: inc. comprehensive manual, P.C. board, five extra components and P. \& P.:-
MONO: $£ 4.90$ STEREO: £9.80

## UNIQUE HYBRID PRE-AMPLIFIER

The HY5 has rapidly established a position in the WORLD as the sole hybrid pre-amplifier to contain all feedback and equalization networks within an integrated preamplifier circuit.

Supplied with the HY5 are two stabilizing capacitors and by the addition of volume, treble and bass potentiometers it is ready for use.

Internally the $K Y 5$ provides equalization for almost every concervable input, the desired function is achieved by use of a multi-way switch or by direct interconnection.

Two distinctive features of the $\mathrm{H} Y 5$ are its inbuilt stabilization circuit, allowing it to be run off any unregulated power supply from 16-25 Volts and a balance circuit which, when linked by a balance control to a second HY5, forms a complete stereo pre-amplifier.

Specifically and critically designed to meet exacting Hi-Fi standards, the HY5 combines extremely low noise with a high overload capability. When used in conjunction with the HY41 and PSU45 forms a completely intergrated system.

## INPUTS

Magnetic Pick-up (within $\pm 1 \mathrm{db}$ RIAA curve) $2 \mathrm{mV} .47 \mathrm{~K} \Omega$
Tape Replay (external components to suit head) $4 \mathrm{mV} .47 \mathrm{~K} \Omega$
Microphone (flat) $10 \mathrm{mV} .47 \mathrm{~K} \Omega$
Ceramic Pick-up (equalized and compen satable) $20-2000 \mathrm{mV}$. variable.
Tuner (flat) 250 mV . $100 \mathrm{~K} \Omega$
Auxiliary 1250 mV . $47 \mathrm{~K} \Omega$
Auxiliary $22-20 \mathrm{mV}$. $100 \mathrm{~K} \Omega$

## OUTPUTS

Main Pre-amp output 500 mV
Direct tape output 120 mV
ACTIVE TONE CONTROLS (Bexendall) Treble $\pm 12 \mathrm{db}$
Bass + 72 db .
INTERNAL STABILIZATION
Enables the HY5 to shat
supply with to share an unregulated
SUPPLY VOLTAGE
$16-25$ volts
PRICE:
MONO: $£ 3.60$
STEREO: $£ 7.20$

## POWER SUPPLY PSU45

The versatile P.S.U. 45 is designed to supply your HY41's +HY5's in stereo or mono format
$\stackrel{\rightharpoonup}{*}$
Specification
Input: 200-240 Volts
Output: $\pm 22.5$ Volts at 2 amps
Overall Dimensions: L. $7^{\prime \prime}$; D. $3.8^{\prime \prime}$, H. $3.1^{\prime \prime}$
PRICE: £4.50 inc. P. \& P.


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| :---: | :---: |
| . $13 \mathrm{in} \times \mathrm{Pin}$ |  |
| plain | 2. |
| 1th tweete | 2.2 |
| twin tweeter | 3-50 |
| Type 350-20 watt witl? |  |
| tweeter 8 ohms. P.P. 37 p | 0 |
| $8 \mathrm{in} \times 5 \mathrm{in} 3,8 \pm 15$ ohms | $1 \cdot 30$ |
| $7 \mathrm{Fin} \times 4 \mathrm{in} 3$ \& 8 ohms | 1.0 |
| FANE 8 in 8 ohm, dual cone | 2-80 |
| CELESTION $\sin 15 \mathrm{obm}$ | 1.50 |
| GOODMANS $10 \mathrm{in} \times 6 \mathrm{in} 3$ |  |
|  8 or 15 ohm |  |
|  |  |
|  |  |
| Eil-form cabinets, teals |  |
| (17in $\times 10 \mathrm{in} \times 6 \mathrm{in}$ ) |  |
| with a 13 in $\times 8$ in or 8 in cut out ( $12 \mathrm{in} \times 12 \mathrm{in}$, 6 in ) witha 8 in $\times$ |  |
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| Add 35p per Cabinet. for post \& packing |  |
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Heavy duty 4-apeed motor with ceparate pick-up arm fitted $\underset{\text { cartridge }}{\text { LP/78 turno ver mono }} \mathbf{5 4 . 5 0}$


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 Sapphire Mono GP91 $21 \cdot 50$; Powerpoint LP/78 80 p .
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 $14 \times 3 \mathrm{in} .16 \mathrm{p} ; 10 \times 7 \mathrm{in} .19 \mathrm{p} ; 12 \times 5 \mathrm{in} .20 \mathrm{p} ; 12 \times 8 \mathrm{in}, 28 \mathrm{p} ;$
$16 \times 6 \mathrm{in} .28 \mathrm{p} ; 14 \times 9 \mathrm{in} .34 \mathrm{p} ; 12 \times 12 \mathrm{in} .40 \mathrm{p} ; 16 \times 10 \mathrm{n} ., 5 \mathrm{p}$,

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Handbook of transistor equivalents

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 \begin{tabular}{ll|ll|ll}
$16 / 450 \mathrm{~V}$ \& 15 p \& $1000 / 50 \mathrm{~V}$ \& 35 p \& $32+32 / 250 \mathrm{~V} .$. \& 18 p <br>
$39 / 450 \mathrm{~V}$ \& 20 p \& $8+8 / 440 \mathrm{~V}$ \& 18 p \& $32+32 / 450 \mathrm{~V} .$. \& 33 p

 

$32 / 450 \mathrm{~V}$ \& 20 p \& $8+8 / 450 \mathrm{~V}$ \& 18 p \& $32+32 / 450 \mathrm{~V}$. \& 33 p <br>
$350+50 / 325 \mathrm{~V}$ \& 50 p

 

$28 / 25 \mathrm{~V}$ \& 10 p \& $8+16 / 450 \mathrm{~V}$ \& 20 p \& $350+50 / 325 \mathrm{~V}$ \& 50 p <br>
$50 / 50 \mathrm{~V}$ \& 10 p \& $18+16 / 450 \mathrm{~V}$ \& 95 p \& $32+82+82 / 350 \mathrm{~V}$ \& 48 p

 

$50 / 50 \mathrm{~V}$ \& 10 p \& $16+16 / 450 \mathrm{~V} 25 \mathrm{p}$ \& $32+82+82 / 350 \mathrm{~V} 48 \mathrm{p}$ <br>
$100 / 25 \mathrm{~V} .$. \& 10 p \& $32+32 / 350 \mathrm{~V} 25 \mathrm{p}$ \& $100+50+50 / 850 \mathrm{~V} 48 \mathrm{p}$
\end{tabular} LOW VOLTAGE ELECTROLYTIC8 $1,2,4,5,8,16,25,30,50,100,200 \mathrm{mF} 15 \mathrm{~V} 10 \mathrm{p}$ $500 \mathrm{mV} 12 \mathrm{~V} 15 \mathrm{p} ; 25 \mathrm{~V} 20 \mathrm{p} ; 50 \mathrm{~V} 30 \mathrm{p}$ $1000 \mathrm{mP} 12 \mathrm{~V} 17 \mathrm{p} ; 25 \mathrm{~V} 35 \mathrm{p} ; 50 \mathrm{~V} 47 \mathrm{p}, 100 \mathrm{~V} 70 \mathrm{p}$ $2500 \mathrm{mP} 50 \mathrm{~V} 62 \mathrm{p} ; 3000 \mathrm{mF} 25 \mathrm{~V} 47 \mathrm{p}$; 50 V 65 p $5000 \mathrm{mF} 8 \mathrm{~V} 25 \mathrm{p} ; 12 \mathrm{~V} 42 \mathrm{p}$; $25 \mathrm{~V} 75 \mathrm{p} ; 35 \mathrm{~V} 85 \mathrm{p} ; 50 \mathrm{~V} 95 \mathrm{p}$.

CERAMIC, 1 pF to 0.01 mF , 4 p . silver Mica 2 to $5000 \mathrm{pF}, 4 \mathrm{p}$. PAPER 350V-0.1 4p, $0.613 p ; 1 \mathrm{mF} 15 \mathrm{p}$; 2mP 150V 15p. S00V-0.001 to $0.054 \mathrm{p} ; 0.15 \mathrm{p} ; 0.258 \mathrm{p} ; 0.4785 \mathrm{p}$. SIL VER, MICA. Close tolerance $1 \%$. 2-2-500pF 8p; 560 $2,200 \mathrm{pF} 10 \mathrm{p} ; 2,700-5,600 \mathrm{pF} 20 \mathrm{p} ; 8,800 \mathrm{pF}-0 \cdot 01$, midd 80 p each. TWIN GANG, "0-0" $208 \mathrm{pF}+17 \mathrm{ppF}$, 65p; 810w motion drive $365 \mathrm{pF}+385 \mathrm{pF}$ with $25 \mathrm{pF}+25 \mathrm{pF}, 60 \mathrm{p}$; 500 pF atend 8HORT WAVE SINGLE. $10 \mathrm{pF}, 80 \mathrm{p}, 25 \mathrm{pF}, 55 \mathrm{p}, 50 \mathrm{pF}, 55 \mathrm{p}$.

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 EEATER TRANS, 6,3 7.3 a . L 2 RRAL 60 p at 2 amp, $3,4,5,6,8,9,10,12,15,18,24$ and 80 F . 22-25 $12 \mathrm{mp}, 6,8,10,12,16,18,20,24,30,86,40,48,60$. 22-25
$2 \mathrm{amp}, ~$
$6,8,10,12,16,18,20,24.30 .36,40,48,60$. 23.25 ( amp.,
amp. $8,8,10,12,18,18,20,24,30,36,40,48,60$.
8 amp. 5, 8 and $13 V$
$5 \mathrm{kmp} 5,8$ and 18 V
$8 \mathrm{amp} .3,5,8,10,18$ and $5-0-5 \mathrm{~V}$
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30 $\begin{array}{ll}1.61 & 30 \\ 2.39 & 36\end{array}$ 2.39
2.62 $\begin{array}{rr}2.62 & 52 \\ 4.39 & 52 \\ 5.80 & 67 \\ 7.77 & 82 \\ 11.20 & \end{array}$ 11.20
20.63 14.10
$\qquad$ Ref. VA WUTO SERIES (NOT ISOLATED) $\begin{array}{ccccccc}\text { No. } & \text { (Wotts) } & 10 & 02 & 11 & 7.3 \times 4.3 & 4.4 \\ 113 & 20 & 11 & 0.115 .210-240 \\ 64 & 75 & 1 & 14 & 7.0 \times 6.4 \times & 600 & 0.115-210-240\end{array}$ $\begin{array}{cccccc}75 & 1 & 14 & 7.0 \times 6.4 \times 6.0 & 0.115-210-240 \\ 150 & 3 & 0 & 8.9 \times 6.4 \times & 7.6 & 0-115-200-220-240 \\ 300 & 6 & 0 & 10.2 \times 10.2 \times 9.5 & .\end{array}$ $\begin{array}{rr}500 & 12 \\ 1000 & 16\end{array}$ $\begin{array}{llll}1500 & 28 & 9 & 13.5 \times 14.9 \times 14.0 \\ 2000 & 40 & 0 & 17.8 \times 16.5\end{array}$ $\begin{array}{llll}1000 & 45 & 0 & 17.8 \times 16.5 \times 21.6 \\ 30 & 17.4 \times 18.1 \times 21.3\end{array}$RMER

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17
115
187
22


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| (1) | 2 zeners. Controls include : bass; ireble; volume; balance; on/aff; "on' indicator; headphone sockets;scratch filter; rumble filter; inputs for magnetic finono; tuncr; dux or tape; tape head etc; Input;selector; nono/stereo switch The construction of this kit was featured in 'Practical Wireless" |  |  |
|  |  |  | ${ }^{31}$ |



For the benefit of those who have a scientific rather than a literary bent, may we explain! Omar Khayyam was a Persian Tent Maker who lived in the llth Century. He was also a Poet, Philosopher and Astronomer. Thanks to the fine translation by Edward Fitzgerald, his best known work is the exquisite "Rubaiyat". It was first produced in England as a pamphlet and offered at sixpence. It didn't sell and was reduced to twopence. As there are only three known copies left in existence you may imagine what they are worth today!

One of the best known verses runs as follows
"Here with a loof of Bread beneath the Bough, A Flask of Wine, a Book of Verse and Thou Beside me singing in the Wildernes. And Wilderness is Paradise enow."

We hope that throws some light on our cartoon. If you read the poem you are bound to come to the conclusion that O.K. liked his tipple-obviously a thoroughly "good type"

What's all this to do with Electronics? You may well ask!!!
Nothing really, but after all this is the Christmas Number, so we should be able to let our hair down. In any case we are sure you must already know that we produce an excellent Components Catalogue which will make interesting holiday reading for you, listing as it does some 6,785 items (over 1,750 illustrated) and costing only 75p including postage and packing (only 55p if you collect it). It also contains 10 vouchers (each worth $5 p$ when used as directed). Are you convinced? No? Well you'd better go out and buy Omar's "Rubaiyat" instead!

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## DOWN TO EARTH

THE obvious sometimes obscures other parts of the picture.

The main purpose of a constructional article is to give all information necessary to enable any suitably experienced constructor to produce for himself a correctly functioning replica of a proven prototype equipment.

Obvious and true; but this is far from all. In fulfilling its purpose a constructional article demonstrates in the most effective and objective manner possible how circuits can be applied in practice. Interesting and unusual applications of the technology are frequently first brought to the attention of the reader through a constructional design. The Biological Amplifier described in this month's issue is a good example of this, since it opens a door to an area of activity perhaps totally unfamiliar to many readers.

Purely theoretical articles are essential and they play their important part in explaining new concepts for circuit design and in suggesting lines for experimentation with new circuit devices. But when it comes to the nitty gritty business of determining precise facts, no amount of theorising can be as convincing, nor as reliable, as the detailed technical description of an actual model, with every component value stated and a performance already established, not merely calculated or predicted.

Amongst our readers there will be a sizeable number who confine their interest in electronics mainly to a study of circuitry and its uses in the abstract-and we suspect they include quite a few who have never held a hot soldering iron in their hand. We believe it would be a great error were they to skip the pages devoted to practical designs. The old saw that an ounce of practice is worth a pound of theory is very relevant even when considering general or non-constructing readers, simply because they do rely exclusively upon the printed word and diagram for knowledge.

Constructional articles help to infuse reality into a subject which, because of the impalpable nature of its workings, is all too liable to be elevated into rarefied regions-with the ready connivance of some of the technical intelligentsia, it has to be admitted. No matter whether we build a particular project or not, careful examination of the text and diagrams will usually teach us something; and certainly help us keep the whole of electronics in true perspective and our own feet firmly on the ground.-F.E.B.

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F. E. BENNETT

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Alinear ohmmeter is an essential piece of equipment for carrying out accurate resistance measurements on thermistors, i.d.r.s, resistive transducers, motor field and armature windings, heating elements, etc. Such an instrument is also extremely useful for checking resistors and resistance combinations, or for batch selection from low tolerance components.

The instrument described in this article is ranged $0-5 \Omega \mathrm{f} . \mathrm{s} . \mathrm{d}$. to $0-100 \mathrm{k} \Omega \mathrm{f} . \mathrm{s} . \mathrm{d}$. in 10 ranges utilising a divide by two facility. These ranges will cover most applications from measuring low resistance motor armature windings to thermistors and transducer tracks.

The circuit is extremely stable but built-in calibration is provided to compensate for drift as a result of battery deterioration or extreme environmental effects. Battery life is six months or more and the indication can be guaranteed to have an accuracy $\pm 1$ per cent of full scale deflection.

## PRINCIPLE OF OPERATION

The principle of operation is well established and is based on the simplified circuit diagram shown in Fig. 1

The resistance to be measured $\left(R_{\star}\right)$ is fed by a constant current generator $\left(I_{k}\right)$ to give an input voltage $\mathrm{V}_{\mathrm{i}}$. This is then amplified and the output voltage $V_{0}$, proportional to the input resistance value $R_{x}$, is measured on a moving coil meter calibrated in ohms. The operational amplifier is connected in its non-inverting mode and has sufficient feedback to ensure good stability of operation, and satisfactory input and output impedance values.


Fig. 1. Simplified circuit diagram of the ohmmeter

## THE COMPLETE CIRCUIT

The circuit diagram of the ohmmeter is given in Fig. 2. The constant current generator TRI is adjusted by VRI, VR2 or VR3 to give the required outputs which may be passed through the test resistance $R_{\mathrm{s}}$ or through the standard range resistors R 1 to R5 by selection of switch Sla or Slb. A fine setting is achieved by adjustment of VR4 to give an accurate full scale meter deflection when the standard resistors (calibrated input) are selected. By scale calibration prior to test readings the accuracy of the system is dependent only on the standard range resistors and the meter.

The operational amplifier ICl is connected to give the required closed loon gain by adjusting the resistor values at the range selection switches $S 2 c$ and $S 2 d$.


Fig. 2. Circuit diagram of the linear onmmeter

The amplifier has a minimum input resistance of 25 megohms and 400 megohms in the unity gain condition so that input resistance ranges of up to 10 megohms may be measured without loss of accuracy. The output resistance is never greater than 10 ohms so a 1 milliamp meter movement may be used without loading the amplifier unduly. The use of such an indicator, used with appropriate series resistors. offers better accuracy and shock resistance as well as being cheaper than a more sensitive movement.

The divide by two switch $S 3$ gives a full scale deflection of 2.5 V when selected and so expands the range of the instrument. The operating conditions as referred to Fig. 1 are shown in Table 1.

## COMPONENTS

Many of the resistors as shown in Fig. 2 are of high tolerance and stability. It is. of course of para-
mount importance to have accurate standard range resistors, as the instrument read-out accuracy is dependent on these.

The current generator is constructed from three presets. a Zener diode and a transistor. The presets may be the Cermet type or the skeleton type, but any silicon pmp transistor and any $5 \cdot 1 \mathrm{~V} 400 \mathrm{~mW}$ Zener will suffice.

The amplifier used in this circuit is preferably the LM741C. but a "A709 may be used with appropriate frequency dependent feedback components. If the latter is used it may be necessary to decouple the supply rails to prevent parasitic oscillation.

The function/on-off switch SI consists of two. unit assembled three-pole wavechange switches with three functions: f.s.d., test and off.

The range switch $S 2$ is made up of four, singlepole 6 -way wafers one waty of each not being used. S3 is a single pole on/off slide switch.

Table 1: OPERATING CONDITIONS

| Switch <br> Position | Resistance <br> Range | $\boldsymbol{V}_{\mathbf{i}}=\left[\boldsymbol{I}_{\mathbf{R}} \boldsymbol{R}_{\mathbf{X}}\right]$ | Amplifier <br> $\mathbf{G a i n}=\boldsymbol{G}$ | $\boldsymbol{R}_{\mathrm{A}}$ | $\boldsymbol{R}_{13}$ | $\boldsymbol{I}_{\mathbf{k}}$ | $\boldsymbol{V}_{\mathbf{0}}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{1}$ | $10 \Omega \Omega$ | 50 mV | 100 | $100 \Omega 2$ | $10 \mathrm{k} \Omega$ | 5 mA | 5 V |
| 2 | $100 \Omega$ | 50 mV | 100 | $100 \Omega$ | $10 \mathrm{k} \Omega 2$ | $500 \mu \mathrm{~A}$ | 5 V |
| 3 | $1 \mathrm{k} \Omega$ | 500 mV | 10 | $1 \mathrm{k} \Omega$ | $9 \mathrm{k} \Omega$ | $500 \mu \mathrm{~A}$ | 5 V |
| 4 | $10 \mathrm{k} \Omega$ | 5 V | 1 | $100 \mathrm{k} \Omega$ | 0 | $500 \mu \mathrm{~A}$ | 5 V |
| 5 | $100 \mathrm{k} \Omega$ | 5 V | 1 | $100 \mathrm{k} \Omega$ | 0 | $50 \mu \mathrm{~A}$ | 5 V |
|  |  |  |  |  |  |  |  |




Fig. 5. Component assembly and wiring details. Wafer switch wiring is shown separate for clarity

```
COMPONENTS
Resistors
\begin{tabular}{llll} 
R1 & \(10 \Omega 1 \%\) & R10 & \(1 \mathrm{k} \Omega 5 \%\) \\
R2 & \(100 \Omega 1 \%\) & R11 & \(10 \mathrm{k} \Omega 2 \%\) \\
R3 & \(1 \mathrm{k} \Omega 1 \%\) & R12 & \(8 \cdot 2 \mathrm{k} \Omega 2 \%\) \\
R4 & \(10 \mathrm{k} \Omega 1 \%\) & R13 & \(820 \Omega 2 \%\) \\
R5 & \(100 \mathrm{k} \Omega 1 \%\) & R14 & \(10 \mathrm{k} \Omega 2 \%\) \\
R6 & \(100 \Omega 2 \%\) & R15 & \(10 \mathrm{k} \Omega 2 \%\) \\
R7 & \(1 \mathrm{k} \Omega 2 \%\) & R16 & \(10 \mathrm{k} \Omega 2 \%\) \\
R8 & \(100 \mathrm{k} \Omega 5 \%\) & R17 & \(10 \mathrm{k} \Omega 2 \%\) \\
R9* & Seetext & R18 & \(2 \cdot 2 \mathrm{k} \Omega 5 \%\)
\end{tabular}
All \(\frac{1}{4}\) watt carbon resistors
Potentiometers
    VR1 \(1 \mathrm{k} \Omega\) vertical preset
    VR2 \(10 \mathrm{k} \Omega\) vertical preset
    VR3 \(100 \mathrm{k} \Omega\) vertical preset
    VR4 \(10 \mathrm{k} \Omega\) carbon linear
Diode
    D1 \(5 \cdot 1 \mathrm{~V} 400 \mathrm{~mW}\) Zener diode
Transistors
    TR1 2N3702
Integrated Circuit
    IC1 LM741C
```


## Switches

```
S1 2-bank, 3-pole, 2-way wavechange switch
S2 Switch kit with four 1-pole, 6 -way wafers
S3 Single pole on/off slide type
Meter ME1 1 mA moving coil meter \(100 \mathrm{~mm} \times 80 \mathrm{~mm}\) (SEW type)
```


## Batteries

```
B1, B2 6 valt style PP1 (2 off)
B3 \(\quad 9\) volt style PP9
```


## Miscellaneous

```
Aluminium sheet 20 s.w.g. \(6 \mathrm{in} \times 4 \mathrm{in}\) and \(15 \frac{1}{2}\) in \(\times 13 \frac{1}{2}\) in
```


## OPERATION

To operate the instrument. S2 is first selected to give the reauired range (if known). The function switch SIb is then pushed to give full scale deflection on the meter which, if necessary, is adjusted by means of VR4.

The unknown resistance maly be connected and the test function switch Sla depressed. Should the ensuing reading be less that half full scale deflection the divide by two facility may be used to give better accuracy of reading.

## PRACTICAL CONSIDERATIONS

In the design stage it was envisaged that leakage currents would not give a significant output reading with both inputs grounded, that is, in the zero input resistance condition. However on test it was found that some i.c.s gave up to 40 , A leakage under maximum gain conditions.

Under these circumstances it will be necessary to introduce a small offset voltage to the inverting input to give zero output deflection. This may be done by adding resistor R9 as shown in Fig. 2.

To calculate this resistance use the equation ats follows, referring to Table $I$ for the appropriate unknown values.

$$
R_{9}=\frac{12 R_{\mathrm{A}} G}{V_{0}}
$$

where $R_{\mathrm{a}}=$ appropriate value to setting (e.g. $R_{\mathrm{a}}=100$ when $G=100$ )
$G=$ gain setting
$V_{n}=$ indicated leakage output voltage $(0-5 \mathrm{~V}$ f.s.d.)

Finally, for low resistance measurements, it must be remembered to take into account lead resistance. This was $0 \cdot 18$ ohms in the prototype unit.


BY FRANK W. H Y DE

## CAMBRIDGE TELESCOPE

Within a few days of operation after the official opening of the Cambridge five kilometres radio telescope at Lords Bridge, new maps were made available.

While the telescope was in construction radio astronomers in America had announced the discovery of radio waves emanating from the bright star Algol. This source has provided the Cambridge workers with the means of calibration much more accurate than hitherto. The radio coordinate system they use is now to be held to a figure better than 0.1 second of arc relative to Algol.

The position of Cygnus $\mathrm{X}-3$ has now been measured to within $0 \cdot 15$ second of are so that there is now a possibility of identifying the X-ray star with an optical object.

The future prospects for this new telescope are very exciting and with its versatility and resolution potential it will bring a great deal of data for analysis.

The resolving of two of the extragalactic objects 3C 86A and 3C 295 gives some idea of the telescope potential, for these two objects had not previously been properly resolved. The object 3C 295 is of particular interest because of its large red shift. The new maps show that this is in fact two units spaced some 4.5 seconds of are apart.

Interest in galactic sources has greatly increased since the molecular lines and infra-red emission from dust clouds have been discovered. One of the new maps using a frequency of 5 GHz has shown, in the contours of NGC 7538, an intense source in addition to the extended emission. This source is a galactic hydrogen cloud and the resolving of the main extended source and the intense source, will enable important deductions to be made.

## PULSARS

The Jodrell Bank programme for the study of pulsars continues to extend the known catalogue. The total is now in excess of 100 pulsars that can be identified. The programme under the leadership of Prof. J. D. Davies added some 18 more to the list during September and October 1972.

Among the new finds was one pulsar that was within $0.6^{\circ}$ of the centre of an extensive nebula known as IC 443. This nebula is a remnant of a supernova which came into being about 65,000 years ago. It is a powerful radio source by reason of the high energy electrons in strong magnetic fields which compose the remnant.
A check on the rotational period of the pulsar showed that it was faster than average yet slowing. The period was 0.334 seconds. Over a period of seven weeks this slowed by 0.33 microseconds. This lengthening of the rotational period is characteristic and is due to the rotational energy of the pulsar being converted into high energy radiation. From the slowing down it is estimated that the pulsar is about 125.000 years old.

The distance from the solar system, based on the dispersion of the radio pulses, would appear to be something of the order of 5,000 light years.
A clue to future methods of search for pulsars is given by the fact that the pulsar is some 50 light years from the point of the original explosion of the supernova. The fact that the pulsars can be thrown out with a large amount of momentum would suggest that search for pulsars should take place outside the supernova remnants as well as within the known area of activity.

## OLD GALAXIES AND QUASARS

There has been an attack on the exponents of the theory that quasars and "new" galaxies can be formed from the exploding of "old" galaxies. There would appear to be some doubt as to whether galaxies are in fact interconnected or that quasars are connected to galaxies.

The 46 metre dish at the Algonquin radio observatory operating at wavelengths of 2.2 and 4.2 cm was not able to detect any interaction. The fact that the radio spectra were normal and no unusual radio effects appeared suggests that there were no violent reactions to provide the birth of a new galaxy much less a quasar. It is also possible that the "bridges" thought to exist between some galaxies may in fact be optical illusions.

## SOVIET PROBE VENUS-8

The results of the findings from the latest Soviet exploration of

Venus indicates that there are similarities between the planet and earth.

The Soviet Academician Vinogradov, already familiar with the analysis of moon rocks, said that the material at the Venus landing site seems to be similar to terrestrial granites. The surface soil contained potassium to 4.0 per cent, uranium 0.0002 per cent, and thorium 0.00065 per cent. These measurements were made by the gamma ray techniques.

Although one spot cannot be taken as representing the whole surface it is likely that the surface has been formed from basic igneous rocks that have undergone metamorphosis. The density of the soil at the landing area was about half that of the earth. The figure obtained for Venus was $1 \cdot 5 \mathrm{~g}$ per $\mathrm{cm}^{3}$.

In the planet's atmosphere there were traces of ammonia (about $0 \cdot 1$ per cent) at the level of the cloud cover. The measurements were made at 33 and 44 Km altitude. The apparatus used was a silica gel column saturated with a sensitive dye for indication. No details of the composition of the cloud cover is available, but measurements of density shows that there is a detectable difference between night and day and that there is evidence that the light from the sun does reach the surface.

## MOON MAGNETISM

Explanations of the residual magnetic field trapped in the rocks brought back from the moon has been examined by various workers. At the moment there are a number of difficulties which prevent a satisfactory explanation to enable a model to be offered for the evolution of the interior of the moon.
To produce the conditions for a magnetic field to have come into being would require that the moon at one time revolved very much faster than earth. This however. would have resulted in its destruction.
There are suggestions regarding the magnetic field which involve certain conditions having been fulfilled during the formation and or the acquisition of the moon. One such theory involves the simultaneous formation of the earth and the moon.
At the present time every hypothesis put forward has faced one or other impossible condition and the field is therefore still wide open. Most of the ideas put forward have involved a liquid core, similar to that of the earth. It could be that this is not a prime requirement and that some other condition existed. The fact that there is no means by which the core of the moon can be examined in detail is the main difficulty.

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## PRACTICAL




This experimental project can demonstrate how neuro systems and electronics can be made to interact and influence certain muscular functions.

Electrical signals from the human body have been used for many years to diagnose certain illnesses and to monitor the condition of a patient during treatment. However, the possibility of using such signals for non-medical applications is a recent development.

The alphaphone, or electroencephalophone, is currently being used by mystics and others to achieve "instant" relaxation. by feeding back brain signals. A similar instrument can also serve as a lie detector monitoring the level of brain signals during interrogation.

Car drivers, athletes, and other people subject to stress, can learn to control their heart rate consciously with a cardiophone, which establishes an external feedback loop between heart and brain.

By D. BOLLEN

The biological pre-amplifier described in this article was designed as a general purpose module for detecting very low frequency signals from electrodes placed on the skin. Also described are a selection of output stages for specific projects, which include the alphaphone, lie detector, and cardiophone.

## SKIN SIGNALS

Brain and nerve cells are similar to logic circuits in that they operate by emitting pulses of electrical energy. Under certain conditions, a group of cells will "fire" together in synchronism, to give a combined output either in the form of a large amplitude pulse (heartbeat) or slow sinusoidal waveform (brain
rhythm). The nerve cells of a tensed muscle emit a stream of random pulses.

As a rough guide to signal levels from a pair of electrodes damped with salt water and placed on *he skin, the chest will yield a heartbeat pulse of, say, 500 microvolts peak, the forehead an deyeblink ${ }^{*}$ pulse of 200 microvolts peak, and the crown of the head an alpha rhythm amplitude of around 50 microvolts r.m.s.

Other brain" rhythös, with the exception of delta, tend to be of low amplitude, generally not more than 10 microvolts, and there is also a rapid, low voltage asynchronous activity when the subject is concentrating or excited.
The very approximate frequency range and psychological characteristics of some brain rhythms are as follows:
ALPHA Frequency $1-4 \mathrm{~Hz}$ for infants, $4-7 \mathrm{~Hz}$ for children and $8-12 \mathrm{~Hz}$ for adults. This frequency can be modified by trtake of certain drugs. Alpha is normally present when closed, and signal is reduced or eliminated by seeing, sleep, unfamiliar sounds, anxiety, and mental concentration. A high level of alpha associated with deep meditation and pain immunity.
BETA Frequency $18-25 \mathrm{~Hz}$. Amplitude increased by anxiety and reduced during the initiation of movements and by muscular activity.
DELTA Frequency $2-3 \mathrm{~Hz}$. Strictly speaking, the name applies only to a type of rhythm associated with brain damage, but signals of similar frequency occur during sleep, accompanied by short bursts or "spindles" at around 14 Hz .
THETA Frequency $4-7 \boldsymbol{H} \boldsymbol{\sim}$ Thought to be related to creative thinking and problem solving activities, and with disappointment and frustration in young people.

## BIO-FEEDBACK

Experiments currently being conducted in the U.S.A. seem to indicate that bio-feedback techniques can be used to alleviate symptoms arising from stress (of which there are many), to reduce heartrate and blood pressure, as well as provide a short-cut to the benefits arising from meditative practices such as Yoga.

The two most convenient ways of feeding signals back to the brain is via the ears or eyes. Fig. 1a shows the aural feedback alphaphone, which picks up subsonic alpha waves from the brain, uses them to modulate an audio signal, and feeds the resulting output to the ears.

The direct link between ear and brain is the feedback path, and the brain itself appears to be capable of adjusting a signal phase to positive or negative. With the alphaphone, the user quickly learns to reinforce his alpha waves, or reduce them, at will.

In the second example of bio-feedback (Fig. 1b), the heartbeat pulse is made to flash a light and the eye feeds this light pulse back to the brain. In time, the subject finds it possible to reduce or increase his heartrate by several beats per minute.
Visual feedback can also be achieved by watching an oscilloscope display, or meter readout of a biological signal.


Fig. 1a. Graphical illustration showing the aural feedback path


Fig. 1b. Graphical illustration showing the visual feedback path


Fig. 2. Graphs of frequency response and noise of the biological pre-amplifier

Table 1. TUNED FILTER


Fig. 3. Circuit of the biological pre-amplifier with component values of $\mathrm{C} 5, \mathrm{C} 6$ and C 7 for different rhythms

## CIRCUIT CONSIDERATIONS

Professional electroencephalographs (brain rhythm recorders) are carefully screened and operated under controlled conditions to achieve a typical noisebandwidth performance of 2 microvolts at 0 to 75 Hz . There is usually some provision for reducing bandwidth by means of switched filters to improve hum and noise rejection.

A noise performance only slightly inferior to that of a professional electroencephalograph is achieved with the biological pre-amplifier described here by sacrificing d.c. coupling. Flicker noise from transistors rises very steeply at sub-audio frequencies, and a considerable reduction in total noise can be obtained by giving the circuit a response roll-off below about 5 Hz .

The pre-amplifier also incorporates a continuously variable bandwidth control which can be set to exclude mains interference in ordinary domestic environments and reduce amplifier noise.

Response and noise curves are shown in Fig. 2. At maximum bandwidth ( -3 dB at 4 and 60 Hz ) noise is less than $2.5 \mu \mathrm{~V}$, while at minimum bandwidth this figure is reduced to less than $0.5 \mu \mathrm{~V}$.

In its basic form, the pre-amplifier is tuned to 10 Hz to give good resolution of alpha signals, but will also handle large amplitude beta, theta, and muscle pulses when set to wideband. However, the circuit can be modified quite easily for narrowband handling of beta, theta, delta, and low frequency alpha signals by altering the values of three capacitors.

All of the circuits given here are battery powered with outputs of less than 6 volts r.m.s., because it is potentially very dangerous to conduct electrode experiments with mains powered equipment. Even
so, a mere 6 volts applied to the scalp is sufficient to cause visual strobing effects, twitching of facial muscles, an unpleasant tingling under the electrodes, and a mild headache if prolonged.

As a general policy, therefore, it is inadvisable to touch any circuit wiring while wearing head electrodes, and the temptation to couple bio-amplifiers to, say, audio amplifiers or flashing light systems, should be strictly avoided.

## PRE-AMPLIFIER CIRCUIT

The circuit of the biological pre-amplifier is shown in Fig. 3. Input transistor TR 1 has a collector current of $50 \mu \mathrm{~A}$ for low noise working with typical skin electrode impedances of around 5 kilohms.

A proportion of TR2 emitter voltage is fed back to TRI base via R3, to set the d.c. operating points of TR1 and TR2. C2 and R4 reduce the gain of the input stage above 100 Hz to combat noise and instability. At 10 Hz , the combined voltage gains of TR1 and TR2 is approximately 300.

A resistance placed between the inverting input and output of an operational amplifier will determine its gain. In the circuit in Fig. 3, an operational amplifier (ICl) uses a twin-T filter between input and output (C5-C7, and R10-R12) to offer a near infinite impedance at the frequency to which it is tuned; a low impedance is presented at other frequencies. Hence, the gain of IC1 is very high at the centre frequency and the amplifier behaves like a high- $Q$ tuned circuit.

The bandwidth of ICI is controlled by a variable resistance VR2 placed in parallel with the twin-T filter. With VR1 at minimum and VR2 at maximum resistance, stage gain is eleven, and can be reduced



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to less than unity when VRI is at maximum resistance.
The values listed for C5-C7 in Table 1 with Fig 3 will give centre frequencies to suit different brain rhythms, but a frequency of 10 Hz will serve for general purpose use. including muscle activity.

Output stage IC2 is another operational amplifier set for a voltage gain of 100 , making the total gain of the pre-amplifier 330,000 . Capacitor C8 has a dual purpose role, firstly as a modulation input whereby the subsonic bio-frequencies can be rendered audible after demodulation by diode DI, and secondly to act as an additional top-cut filter when connected to output 1 . depending on the type of output circuit used with the pre-amplifier.

As it is not practicable to use conventional decoupling techniques with very high gain low frequency circuits. owing to the large values of coupling capacitance required, the pre-amplifier is independently powered by two miniature 9 volt batteries. with output circuits separately powered to avoid interaction.

## CONSTRUCTING THE PRE-AMPLIFIER

Pre-amplifier components can te assembled on a $3 \cdot 7$ in by 1.3 in piece of $0 \cdot 1$ in matrix Veroboard, see Fig. 4. Having cut the circuit board to size, cut the copper strips where shown with a spot face cutter or sharp knife, then proceed to mount and solder components in position.

Take care not to let solder bridge the gaps between copper strips (a miniature soldering iron is essential here) and ensure that capacitor and diode polarities are correctly observed. To complete the panel, solder 3 in lengths of $20 \mathrm{~s} . \mathrm{w} . \mathrm{g}$. tinned copper wire as circuit board leads, in positions numbered 1 to 12 in Fig. 4.

For preliminary testing of the pre-amplifier module, join leads 3 and 4 , and also leads 5 and 6. Spread out the remaining leads so that they do not touch, and connect leads 7,12 , and 11 to two 9 volt batteries. Polarity is shown in Fig. 3.

Fig. 4. Layout and wiring of the pre-amplifier board

## cowponevis.

## PRE-AMPLIFIER

Resistors

| R1 | $330 \mathrm{k} \Omega \frac{1}{2}$ watt metal oxide. | R9 | 10ks |
| :---: | :---: | :---: | :---: |
| R2 | 2.2kS2 | R10 | 33k $\Omega$ |
| R3 | $220 \mathrm{k} \Omega$ | R11 | 33 k S |
| R4 | 10ks | R12 | 15ks |
| R5 | $5.6 \mathrm{k} \Omega$ | R13 | 10ks2 |
| R6 | $470 \Omega$ | R14 | 1MS |
| R7 | $680 \Omega$ | R15 | 22ks |
| R8 | $10 \mathrm{k} \Omega$ | R16 | 10ks2 |

All $10 \% \frac{1}{2} \mathrm{~W}$ carbon except R1

## Potentiometers

VR1 1 MS preset knob trimmer
VR2 100k!2 preset knob trimmer

## Transistors

TR1, TR2 BC109 (2 off)
Integrated circuits
IC1, IC2 741OPA or equivalent (2 off)
Diode
D1 OA47 or 1 B40

## Capacitors

| C 1 | $2.2 \mu \mathrm{~F}$ tantalum 35 V |
| :--- | :--- |
| C 2 | $1,000 \mathrm{pF}$ polystyrene |
| C 3 | $2.2 \mu \mathrm{~F}$ tantalum 35 V |
| C 4 | $100 \mu \mathrm{~F}$ elect. 6 V |
| C5, C6, C7 | 35 V tantalum (see text and Fig. 3) |
| C8 | $0.01 \mu \mathrm{~F}$ polyester |
| C 9 | $2.2 \mu \mathrm{~F}$ tantalum 35 V |

## Miscellaneous

Veroboard 0.1 in matrix, $3.7 \mathrm{in} \times 1.3 \mathrm{in}$. Two 2 mm plugs and sockets.

Now, with a 20 kilohms-per-volt testmeter, measure the voltage relative to lead II of the following: TR1 case ( 1.9 V ), TR2 case (IIV), ICI pin 6 $(9 \mathrm{~V})$, IC2 pin $6(9 \mathrm{~V})$. A variation of about $\pm 0.5 \mathrm{~V}$ can be allowed on the above voltages.

Disconnect the batteries and place the preamplifier on one side for use with the projects to be described next.

## ALPHAPHONE

The purpose of the alphaphone is to apply biofeedback to the brain to increase the level of alpha rhythms and thus promote those "good feelings" which. for some people, are associated with alpha.
Commercial versions of the alphaphone retail in the U.S.A. for $£ 100$ or more, but the version shown in Fig. 5 can probably be built for under $£ 10$ if a pair of headphones are already available.

It is important that the sound heard by the alphaphone user should not be harsh, otherwise this will tend to interfere with relaxation. Another important point is that the alpha signal may have to compete with a background of external, domestic sounds which tend to distract the user and "block" his alpha. Ordinary stereophones, with their soft ear cushions, are excellent for attenuating noise, and a pair may be to hand. Alternatively, cheap mono phones with good sound insulation can be purchased for use with the alphaphone.
Referring back to the pre-amplifier circuit (Fig. 3) if white noise or an audio note is fed to the modulation input (lead 8) it will be mixed with the subsonic alphat signal at the input of IC2, and the intensity of this noise or note will vary in sympathy with the alpha signal after being passed through a non-linear diode DI
In preference to an audio note, which is tedious to listen to for long periods, the alphaphone circuit (Fig. 5) uses a simple white noise generator (D2, R17. R18) to supply a restful "churf-chuff" sound in the headphones, rather like a distant steam locomotive.

An output taken from the pre-amplifier demodulator diode (lead 10) is taken by way of R19 and C10 to a standard type of pre-built audio amplifier. The low alpha frequency is removed by C10 to leave only the higher frequency white noise, which is then amplified and fed to the phones.
The alphaphone fitted to the headband


## LISTENING AMPLIFIER

Almost any small 9 V amplifier, of about 50 mV sensitivity and $\frac{1}{2}$ watt output into 8 ohms, will serve. The author used a Newmarket module PC7+. The value of R17 (Fig. 5) can be adjusted to raise or lower the sound output in the phones, and it may be necessary to reduce the value of C10 with audio amplifiers of high input impedance, to give the required alpha frequency rejection.
The output socket, SK3 in Fig. 5, is shown wired for mono and stereo phones, to drive mono earpieces in parallel, and stereo earpieces in parallel.

The alphaphone should be well screened to minimise interference and instability. A suggested layout, inside a metal box, is included in Fig. 5, and this will give an instrument small enough to fit in a large pocket, and light enough to be carried by a stereophone headband.

The pre-amplifier module can be supported on its stiff wire leads, with a rectangle of s.r.b.p. under the Veroboard copper strips to prevent contact with the base of the metal box.

The PC7+ amplifier can be held in place with two screws, on stand off spacers. VR1 and VR2 may be glued to the front panel with a semi-flexible adhesive, which is easy to prise apart if one of the sub-miniature potentiometers needs to be replaced.

## HEAD ELECTRODES

Electrodes held in place with sticky tape tend to pull at hair and leave adhesive behind on the skin, so the arrangement adopted here was to use spring loaded electrodes mounted on a shaped headband; see Fig. 6.
The electrode configuration shown, with one on the forehead and one on the crown of the head, is suitable for all brain rhythms except beta, and will pick up a good eyeblink pulse. A couple of layers of lint covering the electrode discs will retain the salt water conductive liquid for long periods without attention.

The electrode headband is a strip of ${ }^{3} 6 \mathrm{in}$ sheet Perspex measuring approximately 10 in by $1 \frac{1}{4}$ in and shaped in a steam jet or boiling water to be a loose fit on the top of the head. Holes are drilled at each end of the strip to take the spring loaded electrodes.

The eyeblink pulse is useful to test that equipment is in working order, but it may interfere with brain rhythm experiments, in which case the forehead electrode can be resited near the top of the head by drilling another hole in the Perspex strip. The complete electrode assembly is held in place on the head either with an elastic strap under the chin, or when mounted on a stereophone headband.

Electrode leads should preferably be of twisted or "side-by-side" insulated wire, as ordinary screened cable generates several microvolts of low frequency noise when bent or pulled. Even so, some amount of electrode lead noise is unavoidable if the alphaphone user moves his head around or touches the leads with his hands.

When using the electrodes, the lint covering is first soaked in salt water, and good contact is made with the crown of the head by soaking the patch of intervening hair with salt water. Natural skin grease acts as an insulator and should be removed with hair shampoo or surgical spirit.


Fig. 7. Suggested wiring of the pre-amplifier for an alphameter


## COMPONENTS

ADDITIONAL COMPONENTS FOR ALPHAMETER

## Resistors

R20 $470 \Omega$
R21 $1 \mathrm{k} \Omega$
Both $\pm 10 \% \frac{1}{2} \mathrm{~W}$ carbon
Capacitor
C13 1,000 F elect. 3 V
Meter
ME1 $100 \mu$ A edgewise type level indicator

Switch
S2 Double-pole changeover slide

Batteries
B4, B5 9 volt style PP3
(2 off)
Miscellaneous
Battery connectors
Electrodes (see text)
Metal box (see text)

If mains interference is impossible to avoid, decrease gain (VR1) and reduce bandwidth (VR2) until a satisfactory signal is obtained. The tendency of the pre-amplifier to oscillate at 10 Hz when VR2 is near minimum bandwidth can be checked by blinking the eyes. If there is appreciable ringing after an eyeblink (several cycles of 10 Hz oscillation) increase bandwidth.

## ALPHAMETER

The circuit in Fig. 7 can be variously employed as a relaxation meter, lie detector, and sleep threshold monitor, on the principle that alpha rhythm amplitude varies inversely with anxiety, and is reduced to zero by sleep.

Diode DI in the pre-amplifier module here acts as a meter rectifier, to feed ME1 in Fig. 7 via series resistors R20 and R21, while C13 provides a time constant of 2-3 seconds to smooth out signal fluctuations. C8 in the pre-amplifier module is linked to output 1 (leads 8 and 9) for additional top-cut (see Fig. 2) to reduce interference and unwanted brain noise. Low frequency roll off is also increased by the meter load placed on output capacitor C9 in the pre-amplifier circuit, to remove delta type sleep rhythms.

## CONSTRUCTING AND USING THE ALPHAMETER

If MEl is a miniature edgewise meter, small enough to mount on the front panel, the alphameter can be housed in a metal box of the same dimensions as that used for the alphaphone. A suggested layout is shown in Fig. 7, with ME1 and C13 taking up the space previously occupied by the packaged amplifier.

To use the alphameter as a relaxation meter, set VR1 to maximum gain and VR2 to wideband. Place the electrode assembly on the head, after wet-
ting with salt water, and switch on. When the subject's eyes are open, a reading of about $5 \%$ of full scale should be obtained, assuming that there is little or no mains interference.

Now ask the subject to close his eyes and relax, whereupon the meter reading should rise to about half full scale, indicating a state of moderate relaxation. A gentle tap on a door will cause the meter reading to fall back towards zero, by alerting the subject out of his relaxed state, and the same applies if the subject is asked to do mental arithmetic.

It is interesting to note the increase of meter reading when the subject's eyes are open and be becomes excited, caused by a rise in "brain noise" and theta.

## LIE DETECTION

A similar technique to the above is employed for lie detection, with the virtue that the instrument cannot be misused against the will of the person being interrogated because nervousness or tension obliterates the alpha signal, and renders the lie detector inoperative.

With the subject relaxed and showing a consistently high meter reading, interrogate with a series of simple, undemanding questions, with a short pause allowed between question and answer. If a significant question is suddenly interposed, which calls for a devious answer, the meter reading will fall dramatically towards zero, despite the apparent outward calm of the subject.

To use the alphameter as a sleep threshold detector, the subject is asked to compose himself for sleep, and when the eyes are closed alpha rhythms will initially be generated, causing the meter to read. Just before sleep, the alphameter reading will start to become erratic, and will fall to near zero at the onset of unconsciousness.

Part 2 will describe a brain rhythm frequency meter and cardiophone.

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## BAROMETERS 'N' THINGS

My younger daughter came to me some months ago indignant that her sister ought to be able to know (instinctively) that the weather was going to be awful, particularly as Nanny always reckoned she could tell!

My answer was that her sister probably didn't suffer with rheumatics and that I fancied many people might have this ability, certainly it was no stranger to me. But, I wondered, had anyone bothered to see if there was any correlation between aches and pains, and low (whatever that may mean) barometric pressure.


Fig. 1. Curve shows response vs pain level. Crosses indicate RH.

A relatively diligent search reveals that a high proportion of individuals from almost all ages maintain that they are able to indicate with fair accuracy when the weather is going to take a turn for the worse. However, not content with this (and feeling that relative humidity, RH, might play a hand in it too) ! determined over the next few months to plot a graph (Fig. 1) of pressure and RH versus the subjective level of "ouch", averaged for several people.

It is a particularly interesting fact that although relative humidity doesn't seem to influence things very much, discomfort is definitely related to low pressure, that is pressures much below 1006 millibars (see Fig. 1). Sure, temperature effects have
not been mentioned, but at the time of preparing the graph (during Summer) no great temperature excursions were recorded.

Don't ask me what causes the effect, but there may be some connection with the fact that bone is piezoelectric. This, however, is a subject to which we must return in a later column.

Perhaps in future times, homes will be pressurised so that no one will be able to predict weather conditions, but then they probably won't care either. In the meantime we must either tolerate bouts of "the screws", or go on frequent diving holidays to avoid "weather forecasts"!

## POSITION FIX

Some of the most elegant concepts in engineering are delightfully simple in actuality. No exception to this is an (old, British Rail?) idea recently tested as a feasibility exercise by the U.S. Department of Transportation, New Jersey, to reduce the high costs incurred by existing highway aid systems.

Dear old Britain resorts to motorway telephones spread at intervals of about a third of a mile from one another; but let's face it, this just wouldn't do if our country was even half the size of the United States. The way they have beaten the problem, and simultaneously raised the number of call-points per mile too, has been to adopt the scheme shown simply in Fig. 2.

Initial tests were performed along an 8 mile stretch of "freeway", cables being buried either side of the road. Each cable, which carried two wires, was broken by a currentmonitoring position at four miles, and these stretches were interrupted at intervals of about 200 ft by paralleled push-button switches. At a distance farthest from the monitoring point the cables were terminated in a 100 ohm resistor.

Prior to testing, the line-current (with no buttons pressed) was nulled out. Subsequent operation of any switch thus resulted in a current value corresponding to its position. Hence, a switch near to the monitoring point gave a higher current indication than one operated farther down the cable.

Clearly the whole exercise is based upon a bit of economic thinking, ! just hope that when this system is installed on the large scale no more than one button gets pressed at a time!


Fig. 2.

## ELECTROPHONIC HEARING

What it's called now, I really couldn't say, but back in 1953 a paper bearing the title "Electrophonic Hearing'', and written by Gordon Flottorp of the Psycho-acoustics laboratories at Harvard University, made pretty compelling reading. The effect he discussed, electrophony, was first observed in the early nineteenth century by Volta and others. They found that the sensation of hearing could be evoked by passing low currents through the head. Of course, only d.c. was available at that time, although a.c. effects, more recently investigated, show better promise.

Flottorp reported that in later experiments an "indifferent" electrode was usually connected to the fore-arm while the second electrode could have sites as varied in location as the roof of the mouth or the nape of the neck. Sometimes the tests have been performed using dampened electrodes, while at other times the results have been blessed with greater success when dry ones were employed; actually, under certain conditions, good results were achieved with a moving electrode too.


Quite wide frequency ranges have been obtained; e.g., 200 Hz to 17 kHz , although the intensity does not appear to have been very large. This was typically little more than a whisper at about 30 dB above the threshold of hearing, but, significantly, a discernible signal.

Considerable difficulty has been experienced in all attempts to establish an exact mechanism for electrophonics, since, when passing current directly through the head, the number of possible paths must be very high. It is, therefore, likely that both the 8th cortical (auditory) nerve and the (sensory) organ of Corti are stimulated simultaneously. Indeed, there is a distinct possibility that the skin itself could take the role of a transducer, the sound it produces being sensed as ''direct' hearing!

The voltages employed in these experiments have been kept within the range 1 to 5 volts at currents no greater than 30 mA . Although at higher levels of $V$ and / things are certain to become a little bit excruciating, it does rather look as if the effect ought to be re-examined as a possible aid to the deaf, always assuming that the neural function is intact in the first place.

# 튿TRONIC PIANO 

Part 5

By A. J. Boothman B.Sc.

The final part of the series describes the complete mechanical construction of the piano, together with details on tuning and a number of points concerning the use of the instrument.

## KEYBOARD

Since the commencement of this series of articles F-F keyboards have become available, but as a number of constructors will have purchased C-C Keyboards we now give modification details as promised in the first part of the series.

## MORELLI-A KEYBOARD MODIFICATION

To modify the Morelli-A keyboard to F-F it was necessary to remove the lower $C-E$ section from the main assembly. The wire which acts as a spindle for all the keys was withdrawn from the bottom end by first hammering a nail, whose point was positioned on the end of the spindle, at the top end of the compass. The spindle was then pulled out from the bottom end with a pair of pliers. Prior to this all springs had been released from their bottom clips in order to ease the pressure on the spindle.

Fig. 5.1. The Swedish keyboard with the section C-E separated

In this state all the keys are loose and should be placed on a clean flat surface in the order in which they are removed from the frame. With all keys removed the frame can be cut at the appropriate point with a hacksaw. On completion the keys are replaced in position with the lower C-E section now positioned at the top end.

The wire spindle is replaced from the bottom end, and the final top key (which was originally a C) is now put back to act as the top $F$. This key will require to be made captive since it is free to move up the rod, and for the prototype a brass collet was obtained from a model shop, and clamped on to the end of the rod using the grub screws tapped into the collet.

The frame should now be very carefully stored prior to mounting on the subframe, since the only mechanical linkage between the two parts of the keyboard is the spindle which could easily be accidentally bent.

The author has not attempted to modify the Morelli-B keyboard, but it should be possible to follow a similar procedure to that outlined above.

## KIMBER-ALLEN SWEDISH KEYBOARD MODIFICATION

The Kimber-Allen keyboard is considerably different in construction to the two previously mentioned models, but as in their cases it is not possible to simply rearrange the keys, since the spacing of the fulcrum slots varies over an octave, and it is therefore necessary to cut the frame in order to remove the bottom C-E section and to fix it to the upper end (see photograph).

The first task is to remove some of the keys from the frame by first removing the springs at the rear of the board, which then frees the metal levers which are slotted into the keys. It is then necessary to remove the actuator pegs from the keys by pulling away from the keys. The actuators are a nylon plastic with push fit pegs into two holes in the key. The keys can then be removed.

In order to reduce the risk of damaging any keys, a fairly large area should be cleared-from C to B-flat at the lower end, and B to F at the top end-having first noted the position of $E$ and $F$ keys at the bottom. It would be wise to pencil the key letters on to the alloy frame opposite the relevant actuator holes.

The cut when made will be immediately to the left (viewing the keyboard from the front) of the hole corresponding to note E . This then leaves about half an inch of frame protruding beyond the bottom F key when the assembly is complete, which makes for easy mounting to the wooden sub-frame. For the same reason the line should be drawn to run close to the edge of the hole to leave the maximum amount possible on the piece being cut off, since this part acts as the mounting for the upper end of the keyboard when complete.

Three cuts are necessary, the first being to remove the unpunched section from the bottom end. A line should be drawn to the left of the first (C) hole and a sawcut made keeping the hacksaw vertical. Care should be taken not to damage the fulcrum slot, and to protect the fulcrum knife edge to the left of the slot. The second cut is made in the position described above-i.e. alongside the bottom E hole. Again the saw should be kept as vertical as possible. The third cut is made at the top end of the keyboard and removes the section containing half an actuator hole.

This is necessary in order to finish with the correct spacing between keys. The cut should be made $\frac{1}{8}$ in to the right of the last complete actuator hole, but it is important that the slot in the plastic spacer at the front of the frame is preserved by leaving some plastic protruding beyond the keyboard ( $\frac{1}{4}$ in). When the new section is fitted its plastic can be cut back slightly in order to accommodate the protruding piece. The slot in the plastic is important in that it acts as the front key locator, preventing lateral movement.

The C-C board has two spare actuator holes at the top end which will now take C and D, having swapped the bottom C key with the previous top C key. The new section will accommodate E flat, E and $F$ (the old top $C$ key), with one surplus actuator hole at the top.

## JOINING PLATES

Three steel plates should now be cut and drilled as shown in Fig. 5.2, and mounted as follows:
a. Plate with four holes not countersunk-inside front edge.
b. Plate with four holes countersunk-inside rear edge.
c. Plate with two holes countersunk-outside rear top. This plate should be mounted as far back as possible so as not to interfere with the keyswitch assembly.
Holes should be drilled in the keyboard frame to line up with the metal plates, with the transferred section of frame positioned such that the distance between the two adjacent edges of the actuator holes is $\frac{4}{4} \mathrm{in}$.

If all sawcuts have been made in a square manner there should be a gap between the two sections, however it is unlikely that such a degree of squareness can be achieved with a handsaw, and it may be necessary to make slight adjustments with a file.

The four holes on the front edge of the frame should be countersunk and the screws inserted from the front. The rear edge screws should be inserted from underneath the frame, and the top screws from the top, using the countersink in the plates. When all screws have been tightened the D, E-flat, and E
keys should be tried on the frame in order to check that they do not catch each other, thus requiring further adjustment of the joint.

All keys can then be replaced and the actuators pressed back into position.

## SUB-FRAME FOR MORELLI-A KEYBOARD

The construction of the Morelli-A keyboard subframe, which carries the keyboard, key switch contacts, and terminal strips, is shown in Fig. 5.3.

Assembly of the sub-frame is best carried out using the large baseboard on which to screw down the unit thus ensuring a fully square frame during the process. Particular attention should be paid to this point when fixing the end cheeks which should be both glued and screwed. Using parts obtained from the materials list given the following procedure should be followed.

1. Take the large baseplate and mark out the position for front and rear 2 in $\times$ lin battens, and the end cheeks as shown in Fig. 5.3.
2. Cut the front and rear battens, the end cheeks, the terminal strip mount and the lin $\times \frac{1}{2}$ in hardwood strip as shown in Fig. 5.3.
3. Screw both battens to the baseboard, and glue and screw the end cheeks to the battens.
4. After the glue has been allowed to set remove the frame from the baseboard and fix the $\operatorname{lin} \times$ $\frac{1}{2}$ in hardwood strip and the terminal strip mount.
5. The front of the keyboard frame is fixed to the lin $\times \frac{1}{2}$ in strip with nuts and bolts, whilst woodscrews are used on the rear batten fixing. Holes should be drilled in the metal framework to accommodate the fixings.
Should it later be found that the downward travel of the white keys is restricted by the front 2 in $\times$ lin batten, metal washers should be inserted between the keyboard frame and the lin $\times \frac{1}{2}$ in hardwood strip in order to raise the front of the keyboard. This results in a very slight keyboard tilt which gives a comfortable playing position. Checks on the freedom of the key travel should only be made when the sub-frame is screwed down to the baseboard as the procedure given automatically results in some movement of the hardwood strip as the outer sub-frame is screwed home.

It is recommended that wherever possible the keyboard and sub-frame assembly should be kept on the baseboard whilst the project continues in order to avoid the possibility of mechanical distortion of the assembly, which still depends largely on the base for its mechanical rigidity.

## KIMBER-ALLEN KEYBOARD SUB-FRAME

Fig. 5.4 shows the construction of the sub-frame for the Kimber-Allen keyboard. Again the simplest procedure is to lay out the battens on the main baseboard and glue and screw the four pieces together. The keyboard is fixed to the sub-frame at the ends using long screws passing through $\frac{7}{8}$ in $\times$ $\frac{7}{8}$ in packing pieces. The latter prevent too much stress being applied to the alloy frame when inserting the screws. For additional strength screws can be inserted into the front and rear legs into their respective battens.

Holes of approximately ${ }_{10}^{3}$ in diameter should be drilled at an angle as shown to take the wires from
the keyswitch to the terminal strips on the rear of the assembly. The terminal strips are cut into groups of six, one for each main p.c.b., and depending on the thickness used for the interconnecting wire either one or two holes per strip will be necessary.

The 2 in $\times$ lin planed front bar and side plates are separate from the sub-frame and form part of the main frame which is described in more detail later. Again it is recommended that the keyboard sub-frame assembly should be stored on the baseboard to avoid mechanical distortion occurring.

## THE KEYSWITCH ASSEMBLY

The keyswitch assembly consists of a printed circuit board which anchors gold plated springy wires, and carries the touch sensitivity resistors, and seven rod supports which carry two gold plated rods. Fig. 5.5 shows the arrangement of the various components of the assembly.

Before assembling the parts together, the p.c.b. should be placed on the underside of the keyboard, and used as a template for drilling the $\frac{1}{8}$ in holes required to take the self-tapping screws which fix the switch assembly to the frame.

The front edge of the switch assembly should be placed approximately $\frac{1}{4}$ in from the keyboard actuators, and should be centralised over its length with respect to the actuators. In the case of the Morelli keyboards the p.c.b. pads will roughly line up with the actuators, whilst in the case of the Kimber-Allen keyboard, which is longer, the overall length of the keyswitch is less than the total span of the actuators.

When the gold plated wires are attached they will be bent to fan out towards the extreme ends of the keyswitch, thus resting on the centre of each actuator.

## MATERIAL AND COMPONEWTS FOR REYBOARO SUB-HRAMES AND HAIM CABINEI ASSEMBEY

## SUB-FRAME

Front Batten
Rear Batten
End Cheek (2)
Packing Piece (2)
Mounting Strip
Terminal Mount
Terminal Strips-6 contact
Terminal Strips-2 contact

KIMBER ALLEN
$35 \mathrm{in} \times \frac{7}{8} \mathrm{in} \times \frac{7}{8}$ in
$35 \mathrm{in} \times 1 \frac{7}{8} \mathrm{in} \times \frac{7}{6} \mathrm{in}$
$3 \frac{1}{8}$ in $\times \frac{7}{8} \mathrm{in} \times \frac{7}{8}$ in
$3 \frac{1}{2} \mathrm{in} \times \frac{7}{8} \mathrm{in} \times \frac{7}{8} \mathrm{in}$


MAIN CABINET
Baseboard
Side Pieces (2)
Front Bar
Rear Panel
Rear Panel strip
Top Panel standard version
Top Panel Strip
Mains Panel (3)
Mains Panel Cover
Speaker Panel Support (2)
Switch Panel (2)
Panel Support (2)
Speaker Panel
Panel Stiffners (2)
Main Amplifier Support
Lid Panel
Lid Panel
Music Support Panel
Lid End Pieces
$36 i n$ Piano hinge
2 Carrying handles
4 Cupboard catches
4 24in legs
Covering Material


Fig. 5.5. Keyswitch assembly is screwed to the underside of the keyboard so that the actuators press against the spring wires

## MAKING IT UP

When the holes have been drilled, and the p.c.b. removed from the frame, the keyswitch is assembled as follows:

1. Slide the seven rod supports on to one rod, using the hole nearest the face marked with a " $B$ ".
2. Insert the second rod through the remaining rod holes, and space the supports roughly opposite the holes which have been countersunk from the non-copper side of the p.c.b.
3. Fix the rod supports to the copper side of the p.c.b., using the countersunk screws, nuts and locking washers, with the support face marked " $B$ " closest to the p.c.b.
4. Solder resistors to the pads provided, one on the common positive rail, and the other on the land which also carries the gold plated spring wire. The remaining land is used to anchor the output leads.
5. Solder wire links between the bottom rod and the negative printed track, and the top rod and positive printed track.
6. Cut the gold plated wire into 61 lengths of $1 \frac{3}{3}$ in, and tin approximately $\frac{1}{4}$ in of one end of each wire. Some wire will be left over to provide longer lengths where necessary.
7. Pass each wire between the two rods and solder the tinned end to a main pad. About $\frac{1}{2}$ in should be left to hang over the front edge of the p.c.b. The wire should be pressed flat to the p.c.b. close to the main pad using a match stick, and the soldering iron applied to the wire and land for as short a time as possible.


View of the key contacts at the underside of the keyboard

## MATING WITH THE KEYBOARDS

The assembly can now be fixed to the underside of the keyboard using the self-tapping screws, and a first attempt made to line up each gold wire with the centre of its corresponding actuator.

Some wires will require resoldering to the pads in a different position in order to avoid obstacles such as the rod supports. In certain cases a considerable amount of bending will be required, and it may be necessary to use some of the spare wire to give longer lengths in these instances. The apparently complicated dog leg shapes are perfectly satisfactory in operation.

The final process is to adjust the bend of each wire to ensure that it seats satisfactorily against each rod in turn, whilst gently resting on the actuator in the non-depressed state. As a further precaution to avoid the wires springing off the actuators, small slots can be filed in the centre of each actuator to prevent the wires sliding.

## WIRING THE SUB-FRAME

To complete the keyboard sub-frame assembly wires must be taken from each switch pad to the corresponding terminal strip at the back of the assembly. The pads are labelled with the note concerned and as suggested earlier one or two holes have been drilled against each terminal strip which carries all octaves of a particular note.

A single link is also required between the sixth contact on each of the terminal strips, as this acts as the common sustain connection. Wires are also taken from the positive and negative pads on the keyswitch p.c.b. to a two contact terminal strip mounted on the left hand (power supply) end of the keyboard sub-frame.

## CABINET ASSEMBLY

The general layout of the cabinet is shown in Figs. 5.2 to 5.9.
The cabinet construction has been kept deliberately simple, whilst achieving some aesthetic quality.



Fig. 5.8


Fig. 5.3. Baseboard layout end view of sub-frame ase

Fig. 5.4. Baseboard layout and end view of sub-frame Fig. 5.6. General assembil panel dimension in paren keyboard only

Fig. 5.7. End pieces for the board. Two are required. endpiece (mains input). endpiece

Fig. 5.8. Mains input pane from $\frac{1}{1}$ in plywood

Fig. 5.9. Keyboard control panel is shown



Fig. 5.6
or Kimber-Allen keyboard
for Morelli-A keyboard and mbly
for Kimber-Allen keyboard
, of the cabinet. |The top hesis is for the Morelli-A
main box using $\frac{3}{4}$ in block(1) Cut-out for left-hand
(2) Cut-out for right-hand and guard pieces made panel details. The left-hand


(a) Envelope input circuit as used during the initial setting up procedure on the pitch board

Fron touch
control
$0-$

(b) Modified envelope input circuit to give touch sensitivity. R45A is mounted on the keyswitch assembly as shown in Fig. 5.5

Fig. 5.10. Circuit changes to give individual key touch sensitivity

The two end pieces are permanently fixed to the baseboard, joined by the front bar. All other parts are detachable in order to give easy access to the various sub units.

Two end pieces should be cut from $\frac{3}{3}$ in block board, each one requiring a different cut out as shown in Fig. 5.7. The position of the speaker panel supports is also indicated in this diagram. The two pieces should be glued and screwed to the chipboard base, with the front bar positioned $\frac{1}{2}$ in from the front edge of the base board.
The mains input panel should be cut and drilled from $\frac{1}{n}$ in plywood as shown in Fig. 5.8. The other three pieces shown form a box which covers the back of the mains switch, etc. to avoid the danger of electric shock.
The speaker panel was covered earlier, and is a removable unit which seats on the supports shown.

The rear panel has a $\frac{1}{2}$ in $\times \frac{1}{2}$ in strip glued along its top edge which acts as a retainer for the top panel. The rear panel is screwed to the end pieces and is easily removable.

The top panel is retained in position by the strip attached to the rear panel, and by spring clips positioned at the front, with the female part screwed to the end pieces. The clips used were of the common cupboard door type and are shown in Fig. 5.7. The slots cut in the top panel allow heat to be released from the power supply and main amplifier, whilst the $\frac{3}{5}$ in $\times \frac{3}{n}$ in strip acts as location for the lid when used in the music stand position.

Dimension W is dependant on the keyboard used, and if the constructor uses the Morelli-A keyboard he may like to add a $\frac{1}{4}$ in strip on the sloping edges of the end pieces in order to retain the same styling as achieved when the other keyboards are used.

## CONTROL SWITCH PANELS

Fig. 5.9 shows details of the construction of the keyboard level switch panels. Dimensions X and $Y$ depend on which keyboard is used, and dimension $Z$ from Fig. 5.7 is also given in the table. The 2in $\times \frac{1}{2}$ in $\times \frac{1}{2}$ in support strips should be glued to the side pieces. Using the dimensions given it should be possible to line up the front edge of the keys, when fitted on the keyboard sub-frame assembly, with the front edge of the keyboard switch panels.

## COVERING THE CABINET

The choice of finish to use on the cabinet greatly depends on finances. To give maximum wear and a smart appearance a heavy grade of black p.v.c. sheeting was used on the prototype, with speaker fabric covering the whole of the speaker panel. Other alternatives could be a Contact/Fablon type material, or paint.

## TOUCH SENSITIVITY

Individual touch sensitivity is achieved on each key by the use of a keyswitch assembly which is effectively 61 single pole changeover switches. The two contacts are actually busbars, common to all switches, one at ground and the other at the attack potential. See Fig. 5.10.

Referring to Fig. 5.10b the keyswitch is normally in the grounded busbar position. When a key is depressed the switch leaves the grounded busbar, thus allowing C28 to charge via R45A. the bottom end of which had previously been shorted to ground. The attack level passed on to C33 is determined by the atiack voltage from the touch control, reduced by the voltage across C28.

The attick voltage is applied when the second busbar is reached, and the voltage across C28 increases as the time taken to traverse between the two busbars increases. The attack level passed on to C33 is therefore reduced as the time to traverse increases, and is therefore proportional to the average speed of key travel after it is depressed.

This is not of course exactly the same mode of touch sensitivity as on a conventional piano, where the final speed of key travel is the important parameter, but this system does give a degree of realism. The range of touch is considerably less than on a string instrument, but both classical and pop pianists have demonstrated a noticeable level of expression in their playing of the second prototype, relying solely on the keyboard touch sensitivity without access to the soft pedal.

In order to bring the touch sensitivity into operation the value of R54 is altered and R45 removed from the pitch board. R45A is mounted on the keyswitch assembly as described earlier.

Some constructors may find it interesting to experiment with the values of R44, R45A and R54 in order to vary the dynamic range of the touch sensitivity.

## TREMOLO PRESET DEPTH

With all units interconnected the preset tremolo depth control (VR6 in Fig. 2.5) can be set. With VR2 in the maximum position and the tremolo on, adjustment of VR6 will produce zero tremolo effect at one extreme, and reduce the signal output to zero at the other extreme. A position can be found where maximum tremolo swing is obtained.

## TUNING

Tuning can only be carried out by someone with the so called 'musical ear', so if you do not have one you will need to involve a friend or professional tuner.
Commence with A (the first above middle C) using a tuning fork A440. Strike the tuning fork against the cabinet, and use the top panel as a sound board. The pitch board test jig is used to hold the board under alignment, first using the capacitor values suggested in the table, and then small value changes made until A can be achieved with the inductor setting approximately in mid position.

The ' A ' board is then put into the pitch module frame and a second board put in the test jig. The author preferred to tune the $F$ board next by comparison with the ' $A$ ' and thus listening to a major third interval. Again the values given for ' $F$ ' in Table I (last month) are tried first with modifications as necessary.

When the ' $F$ ' board seems satisfactory it is inserted into the module, and the ' C ' board put into the test jig. The major chord F-A-C is easy to recognise and adjustments to both ' $C$ ' and ' $F$ ' (now in the module) can be made to give a satisfactory sound. ' $A$ ' should not be adjusted, but can be checked a number of times against the tuning fork in order to satisfy the constructor that he has tuned it correctly.

Using "C' as the base, ' $E$ ' can now be tuned (major third) followed by ' $G$ ' in the major chord C-E-G.

Using ' $G$ ' as the base, ' $B$ ' can be tuned (major third) followed by ' $D$ ' in the major chord G-B-D, thus completing the white notes.

Two approaches are now possible to tune Effat. Some would find evaluation of the F7 chord F-A-C-Eflat to be relatively easy, whilst others may prefer to use the major third B to Dsharp (Eflat)that is using B as the root.

Bflat can be tuned from the major chord Eflat-G-Bflat.

A choice exists again to tune Aflat, either using the Bflat seventh chord Bfat-D-F-Aflat or the major third E-Gsharp (Aflat). The full major chord E-Gsharp-B can be used to make judgement easier.

Fsharp can be tuned using the major third interval D-Fsharp, with A completing the major chord.
Csharp can be tuned using the A major chord A-Csharp-E.

The cycle may need repeating a number of times, and other chords will probably produce some out of tune effects which small adjustments in tuning will bring into line. Tuning the piano should not be any more difficult than tuning a guitar, but it will take longer, and requires patience.

## FINAL ASSESSMENT

The piano can now be played, and an evaluation made by the constructor of the properties of his instrument. Probably the most important observation will be an assessment of the relative outputs
from each note on the keyboard using a constant playing action. If the constructor is dissatisfied he can adjust individual notes or complete pitch boards for increased or reduced output by variation of the resistors quoted in an earlier part of the series.

## USING THE PIANO

Whilst the accent in the early part of the series was on constructing an instrument to use in the home, the electronic piano has very great potential for use in modern groups or bands, particularly since the pitch of pianos normally present in private rooms in hotels is often very flat, which can-cause very great difficulties for wind instrumentalists in the same band.

The piano can be connected directly to a public address or guitar amplifier thus offering the pianist power capabilities much higher than he has known before. In this connection it is important that very heavy duty speakers are always used when high amplification is involved.

The author uses an amplifier rated at 25 watts r.m.s. driving a speaker with a nominal rating of 50 watts (Fane POP-50). This has withstood a considerable amount of hard work with the piano and three microphones on the same system.

Rating of loudspeakers in this application can be very misleading, since music power is the important factor, and a 25 watt r.m.s. amplifier is capable of driving 50 watts of music power thus requiring a suitably rated loudspeaker.

## playing the piano

To close some comments are now made on the technique of playing the piano.

The normal switch setting when using the touch sensitive system is maximum attack. The other attack positions can be used to give a less percussive characteristic for variety.

It is vitally important that the use of the keyboard level control is fully understood. Five positions are offered, but position one must only be used for single note solos, since the power output in this position is the same as a five note chord in position two.

Considerable distortion will occur if this is ignored.

With position two selected, the volume control should be set to give the maximum required listening level. Selection of level for any particular passage is then controlled by the keyboard switch and the foot pedal. The volume control is not varied during a performance.

The natural tendency when trying to achieve greater attack on a piano is to hit the keys harder, up to a stage where great physical effort is involved. This tendency should be resisted on the electronic piano where as mentioned earlier the attack function is different.

It will be found after practice that the touch can be very light, leading in some cases to an easier playing technique, whilst retaining a range of expression.

Note: There are a number of resistor value corrections in Fig. 3.1. These are: R84 should be $100 \mathrm{k} \Omega$, R85-R88, $56 \mathrm{k} \Omega$ ( 4 off) and R110, 270S2. Also, in Fig. 3.4, TR16 collector, base and emitter are labelled in reverse.

# $\square \square \square$ <br> A PROPOSED NEW BROADCASTING SERVICE 

IMAGINE sitting down in front of your television set, and at the touch of a switch seeing the latest news, up to the minute stockmarket prices or a list of the day's television programmes with a summary of each. Just a pipedream? Not any more, for the BBC has developed a system which can allow you to do just that, and demonstrations should be underway as early as the middle of 1973.

## THE CEEFAX SYSTEM

The system is known as Ceefax and the form that it will take is shown in the photograph. Fig. 1 shows how the Ceefax system fits into the existing television transmission network. At the transmitting end it will mean incorporating a terminal into the broadcasting system so that the pages to be sent out are mixed with the normal picture signals. This will probably mean using the frame synchronisation pulses on which are superimposed the digital signals.

At the receiving end the user will have to purchase a storage and display unit which will probably have a storage capacity of up to 30 pages. The general block diagram of the terminal is shown in Fig. 2.

The user will type out whichever of the 30 pages he requires and it will immediately appear on the screen, in a teleprinter type format, either on a blank background or superimposed on the normal television picture.

## COST TO THE CONSUMER

The cost of the additional electronics to the domestic user will undoubtedly decrease as technology develops but it is estimated that in quantity production the cost will be approximately the same as a monochrome set or a tape recorder.

Any of the pages stored in the receiving unit can be updated in a matter of seconds, and some pages could be updated every few seconds to present information in book form or perhaps to provide subtitles for television programmes in foreign languages when we enter the Common Market.

## INFORMATION AVAILABLE

Examples of the information which can be transmitted using the Ceefax system are limitless but some of the more obvious are: specialised weather
$!$


Fig. 2. The domestic receiver uses an addl-on decoder and store ready for selected display on the screen

Fig. 1. The Ceefax programme material is injected into the normal television transmission in digital form
forecasts; a list of the day's television programmes with summaries; specialised programme news for the coming week; facts and figures relevant to current affairs programmes; public notices; up to the minute news (with the option of a flash service whereby important events are flashed onto the screen over the normal programmes), commercial news such as stockmarket prices; sports information; and entertainment in the form of recipes or crossword puzzles.

In the future it is hoped to co-ordinate Ceefax with the Post Office so that pages of information can be received over the normal telephone lines. This will bring the television set into the realms of the computer terminal and no doubt eventually personal facts stored in a computer, such as bank statements, will be at the fingertips of the user through the Ceefax system.

## OTHER POSSIBILITIES

There is also no reason why educational institutions should not use one of the avalable channels for passing information on to the students, in reply to queries made over the phone. The telephone system could also be useful for receiving telegrams and telex messages, there being no need for the recipient to be present when the niessage is sent.

Although the Ceefax system seems at first like a simple development it is one of those advances. like the introduction of the telephone system, which could hatve a profound effect upor all our lives as it brings centres of information within easy reach of everyone. It could easily mean the end of newspapers as we know them today for they would be unable to provide such an immediate source of news and would probably only serve to give deeper insight into the news.
The first trials will probably use equipment manufactured by the BBC, but industry will have the opportunity of producing Ceefax units for a public service operation.


## NEN

 from Goodmans for constructors
## Din 20 Kit

20 watt, high fidelity loudspeaker kit contains all parts necessary to complete the system, except timber and other material for the cabinet itself, with detailed, illustrated instructions. Specification: 20 Watts DIN 4 ohms impedance, 8 ins bass unit, dome HF radiator, crossover frequency $4,000 \mathrm{~Hz}$.


Axent 100


Dome HF Radiator with integral cross over. Capable of high frequency sound reproduction with negligible distortion in systems rated up to 30 Watts DIN, this 'state of the art' drive unit has an integral crossover which cuts frequencies below 3 kHz at a rate of 12 dB /Octave.

## Audiom 100

12 inch high fidelity bass loudspeaker.
For use as a bass unit in two-way systems, the sensitivity and high frequency roll-off of the Audiom 100 has been tailored to match the Axent 100.


## Goodmans

 Sound reasoning.
## THOAN

[^3]Look out for the name Akerman. You'll hear a lot about him and a loi from him starting on January 1 , 1973. For that's the date that Jack Akerman takes up his new role as managing director of Mullard Ltd. in succession to Dr. F. E. Jones who leaves to join the Board of Philips Industries Ltd.
Akerman is no newcomer to Mullard-he joined the company in 1936--but it is only in recent years that he has become prominent. Five years ago he moved into the higher echelons through his appointment to the Board of Associated Semiconductor Manufacturers Ltd., an assosiated company of Mullard. In 1969 he became head of Mullard's consumer electronics division and a year later, in 1970, he became commercial director and a member of the Mullard Board.


Dr. F. E. Jones came to wide public notice through the celebrated "Brain Drain" report which came out during the Wilson administration. But he was well known in the electronics industry Iona before that both as a brilliant engineer and businessman-a rare combination. He joined Mullard from the Royal Aircraft Establishment, Farnborough, where he was deputy director. After six vears as technical director he became MD in 1962 so he has had a full ten years of the top job during which major expansion has been almost a routine.

Jack Akerman comes to the top job at an interesting time. Business is on the upturn and it is not only the consumer component side of Mullard business that is now doina well after the recession. Professional components are also booming and delivery times are lengthening.

## LED PRICE BATTLE

Last month | mentioned the bright outlook for optoelectronic devices. Since then I have met Mike Meltzer, in London for a few days representing Monsanto interests in light-emitting diodes and associated devices.
Monsanto announced price cuts of typiaally 30 per cent but in a few cases up to more than 40 per cent on standard product lines. Not to be outdone, Sperry announced in the U.S.A. cuts of up to 40 per cent on some of their devices and Litronix, another US company, has also joined in the price battle.

How far prices will tumble is anybody's guess but most companies in the LED business now believe that the time is ripe to go for a huge volume market. Monsanto's Meltzer told me that if pocket calculators fall again in price next year, as they are likely to do, then that one single market could generate a demand for 100 million digits a year.

A real enthusiast on his subject, Meltzer spoke of the untold millions of digits which will be needed for as yet untapped markets like cars and the telephone industry. One of Meltzer's forecasts is that by 1980 the LED industry will be as large as the whole semiconductor industry today. Nice work for the survivors after the price war has taken its toll.

## NEW FIELDS

Ever anxious to tap new markets. the industry is queuing up to get orders from oil companies. These new rigs scattered round the North Sea are becoming big business. Not so much a drilling platform, more a small town by the sound of things, and there are now over 40 of them all needing not only high grade communications systems but also beacons and other devices for bringing in helicopters as well as surface vessels.

One company, Marconi Marine is doing well supplying complete packaged radio stations including radio officers. Other companies tend to supply individual pieces of equipment.

## the valveis not dead

Biggest ouptut in the U.K. of any single component group is still in the field of domestic receiving valves and tubes which are running at a rate of $£ 60$ million per year. Industrial valves and tubes are worth another $£ 25$ million per vear. Despite the advent of solid state, valve and tube manufacturers are still doing excellent business.

If you think there is nothing new in the pipeline I suggest you
remember the initials CFA and FWI of which you may hear more in the next few years. CFA stand for Crossed Field Amplifier and FWI stands for Fast-Wave Interaction. CFA and FWI valves are reported under development at Enalish Electric Valve Company.

The CFA combines the advantages of the magnetron and the travelling wave tube and could open up a new range of advanced radar systems. The FWI could open the way to much higher microwave power.

## PRODUCTION SHOW BREAKS RECORDS

The fifth Internepcon Show at Brighton again broke all records. It concentrates almost entirely on packaging and production lechniques for the electronics industry and is supported by a full-scale conference. This year it attracted over 14,500 visitors including a large number from overseas, some 1,000 of these having registered on the first day.

The Nepcon shows are popular world wide and are money spinners for the organiser as well as the exhibitors who seem so well pleased with results that they come back every year for more despite rising costs.

January sees the second Internepcon held in Tokyo. Ten British companies will be exhibiting. Then next June, the organisers will be launching the first European edition in Brussels which will be called Internepcon Europe 73. Sixty British companies are expected to be there as one of the national groups. This will be followed in October by the usual show in Brighton for which many of this year's exhibitors have already booked space.

## EXPANSION TARGET

Semicomps Ltd., the component distributors recently acquired by Simms Motor Division of Joseph Lucas, is embarking on a five-vear plan to expand its business to $£ 8$ million a year turnover. Managina director Bill Richardson, founder of Semicomps, is confident that the new financial backing he now enjoys will ensure he attains his obiective.
The big turnover will be achieved by setting up extra companies to give complete U.K. coverage. The parent company, based in the South East, already has a subsidiary in the North and another in Ireland. More will be set up and each will be expected to contribute $£ 750,000$ to $£ 1$ million turnover to the group total. First two areas to see new Semicomps companies will be the Midlands and Central Southern England.

## JUST PUBLISHED '73 AUDIO-TRONICS

The great new 1973 edition of Lasky's famous Audio-Tronics catalogue is now available-FREE on request. The 48 newspaper size pages-many in
full colour-are packed with 1,000 'slof items from the largest stocks in Great Britain of everything for the Radio and Hi-Fienthusiast, Electronics hobbyist Serviceman and Communication's Ham. Over hall the pages are devoted exclusively to every aspec of Hi-Fi (including Lasky's budget Stereo Systems and Package Deals). Tape recording and Audio Accessories, see the great new Lasky's Credit Plan Scheme-enabling you to buy your ideal choice o equipment on easy rerms. Sed your name and regular mailing list

## LEAK TEAK CASES

Teak or Walnut case for Stereofetic tuner only. LIST PRICE LASKY'S $\mathbf{2 2 . 5 0 \quad \text { C \& }}$ 25p C7.37 PRICE
Double case to hold Stereo 30 or Stereo 70 and Stereofetic tuner in Teak or Wainut
LIST PRICE LASKY'S 44.95
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PRICE

## LEAK TRUSPEED

2-speed ${ }^{45}$ and ${ }^{33 \frac{1}{2}}$ rpm complete with plinth, cover IST PRICE 669.50



## SINCLAIR PHASE LOCK LOOP STEREO FM TUNER

Incorporates varicap diodes, printed cir cuit, coils, squelch
circuit I.C. Decoder etc. supplied com pletely built and tested and ready to be mounted into any cabinet you choose. It may requirements $2 \mathrm{~S} / 30 \mathrm{~V}$ DC
LIST PRICE LASKY'S 45.75
$£ 25.00$
PRICE
:SR TAPE DECK


FANTASTIC VALUE, ONLY Lasky's can offer you a tape deck at such an amazing price The BSR TD2 tape deck operates by a simpl reliable mechanism using the minimum of controls.
With $\frac{1}{2}$ track mono heads. Incorporates fast wind and fast rewind, records at $3 \frac{3}{2}$ ips. giving up to 3 hrs playing time, takes up to 5 zin spools. plate, lying above front
ALSO SUITABLE FOR USE AS A TAPE TRANSPORTER
LASKY'S $\& 8 \cdot 95$
PRICE

BELTEK C5700

## 8 TRACK

Stereo c
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## INTEGRATED CIRCUITS

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IN many applications it is necessary to drive TTL circuits from a non-TTL input. How this may be achieved will depend on the input signal, but one possible arrangement is shown in Fig. 4.1. This is suitable for many input waveforms and there is no danger of the logic level for the integrated circuit being exceeded.

When the transistor is turned on its collector is pulled down to the logical 0 level for the gate, producing a 1 at the output. This circuit can be used where the rise and fall times of the input are faster than $1 / \mathrm{s}$. Unfortunately when high speed gates are driven directly by input signals which have a slow rise or fall time it is possible for oscillations to occur while the input signal passes through the linear region of the input/output characteristic.

The gate acts as a linear amplifier with a high gain, and various feed back paths exist which contribute to instability and consequent false outputs.

One solution for slow rise time inputs is to use a Schmitt trigger. The basic Schmitt trigger has a positive feedback arrangement which introduces hysteresis to the input/output characteristic, and its snap action prevents oscillation.

## SCHMITT TRIGGER USING GATES

A Schmitt trigger can be constructed by using gates as in Fig. 4.2, but this circuit is not temperature stable, needs to be driven from a low impedance source, and requires discrete components.

## THE SN7413

The SN7413 consists of two identical Schmitt trigger circuits, similar to the usual transistor arrangement. Each functions as a four input nand gate with different input threshold levels for positive and negative going signals. This hysteresis is typically 800 mV . The SN7413 is fully compatible with the other TTL circuits in the range.

Propagation delay from input to output is typically 16 ns. The package details and parameters are shown in Table 4.1. It has built-in temperature compensation to maintain high stability of the threshold levels


HYSTERISIS Having a different switching threshold for increasing and decreasing signals.
INTERFACE A circuit or "black box" which joins dissimilar systems.


Fig. 4.1. One arrangement for driving logic from a non-TTL source
and can be triggered from the slowest of input waveforms and still give a jitter-free.output.

## PULSE SHAPER

The SN7413 can be used as a pulse shaper interface between slow waveforms and fast TTL circuits as shown in Fig. 4.3a and b.

## SINE TO SQUARE WAVE CONVERSION

This is perhaps one of the simplest applications of the Schmitt trigger, and is shown in Fig. 4.4. The input of the circuit is biased midway between the upper and lower threshold points by R1 and R2 which gives a 50 per cent output duty cycle for sinusoidal inputs. The circuit can also be operated in a self biased mode. Fig. 4.5. These circuits can be used up to 5 MHz or more.

Although the input impedance is not high these circuits can usually be driven directly from an oscillator: care should be taken that the input logic levels are not exceeded.

## R-C MULTIVIBRATOR

The circuit of Fig. 4.6 can be used over a wide frequency range for experimental work with TTL circuits, and makes a reliable oscillator from 1 Hz to 1 MHz by selection of the capacitor value: Suppose that initially the capacitor is discharged and the output is at logical 1. The capacitor then charges towards $V_{c,}$ through resistor RP until the upper threshold voltage is reached. The output then changes to a logical 0 and the capacitor discharges through R1 to the lower threshold voltage, and the cycle repeats.
The duty cycle is $2: 1$ rather than $1: 1$ because the internal 4 kilohms resistor on the base of the input multi-emitter transistor acts as a current source. The capacitor required for a given frequency is best found empirically, typical values are shown in Fig. 4.7. This circuit only has a fan-out of two and should be buffered by the other Schmitt trigger in the package or by a SN 7400 if a larger fan-out is required.


Fig. 4.2. A Schmitt trigger using gates and discrete components


Fig. 4.3. Arrangements using the SN7413 for driving logic from a non-TTL source


Fig. 4.4. Sine to square wave conversion using the SN7413


Fig. 4.5. Sine to square wave conversion using the SN7413 in a self biased mode


Fig. 4.6. The SN7413 as a multivibrator


Fig. 4.7. The capacitor values required for various frequencies of oscillation in Fig. 4.6


Fig. 4.8. The circuit in Fig. 4.6 modified for an approximately 1:1 mark to space ratio

output


Fig. 4.9. The SN7413 used as a gated multivibrator

The circuit can be modified for an approximate 1 : 1 mark to space ratio as in Fig. 4.8, but the best way to obtain an exact 1 : 1 mark to space ratio is to generate twice the required frequency and divide by two.

## GATED OSCILLATOR

The basic multivibrator can be gated by adding an SN7401 open collector Nand gate, Fig. 4.9. The fredback path from the output of the SN7413 to mre input of the SN7401 prevents the gate signal from acting until the output of SN7413 is at a logical 1, so that the oscillator will always produce a complete number of cycles.

When the oscillator is released the capacitor has to charge up from ground to the upper threshold level of the Schmitt trigger, so that there is a delay between the negative going edge of the start pulse and the beginning of the output pulse train. This delay is very approximately 0.7 of the total period of oscillation.

## CONCLUSION

This article has shown several applications of the SN7413 Schmitt trigger, but it is especially useful as a straightforward wide range oscillator for future experiments.

## Next month: Operational amplifiers with power stages

I.C. LINEAR OHMMETER
continued from page 29


Underside of ohmmeter control panel. Note the strapped battery pack assembly

## CONSTRUCTION

The case is made from 22 s.w.g. aluminium and cut out as shown in Fig. 3. The front panel is cut out and drilled from 20 s.w.g. aluminium sheet.

A piece of paxolin. $2 \frac{1}{2}$ in by $3 \frac{1}{2} \mathrm{in}$ is drilled according to Fig. 4. Wiring pins should be added to the board's extremities as seen in Fig. 5 and components assembled and inter-wired according to the dotted lines.

All control panel components should now be mounted, which includes the wafer switch assembly S2.

The component board is bolted onto the meter terminals and connections from these to the appropriate board pins made. Final wiring should be carried out according to Fig. 5.

To make up the power supplies a battery pack consisting of one 9 V and two 6 V batteries are strapped together (see photograph) and housed in the case.

## SETTING UP

The resistance settings of VR1, VR2 and VR3 to give the required current generator outputs are approximately $900 \Omega 2,9 \mathrm{k} \Omega 2$ and $90 \mathrm{k} \Omega 2$ respectively These may be roughly set before switching on. Fine adjustment may then be carried out with the instrument on, to give full scale deflection.
The output should, of course, be perfectly linear but this may be checked by testing with some high accuracy resistors.

# PRTENTE <br> RIECISW。 

LIFE TESTING

The not particularly cheerful but daily more pressing question of a definite test for life applicable to medical transplant techniques is raised by $E$. Roy John of New York in BP 1257 114. Obviously the need for a definitive "absence of life" test is crucial in transplant surgery. If removal of the transplant organ is delayed unnecessarily, the organ will be useless. If the organ is removed too early, the doctor will be acting immorally and illegally.

The original test for life was the existence of breath and heartbeat, but this is clearly outdated. Attempts are being made to use electroencephalograph and electrocardiograph readouts for sensing brain and heart activity respectively. But amplifier noise can be mistaken for brain activity-or more important, it may be mistakenly assumed that what is in fact brain activity, is simply amplifier noise.

The most recent technique is to stimulate the brain and detect response; but to avoid false results, due to areas of the brain being damaged, three of the major sensory systems have to be tested. All this can become very time consuming and time is of the essence in transplaṇt surgery.

The patent suggests using digital and analogue systems for evaluating the $E E G$ and $E C G$ readouts after visual, auditory and somatic stimuli. To this end a graphic recorder is used which has at least three recording channels. An EEG has its input electrode connected to the patient and its output con-
nected to the first of these channels. An ECG has its input electrode connected to the patient and its output connected to the second of the channels.

Finally, a data averaging device has its input connected to the outputs of both the EEG and ECG and its output connected to the third channel of the araphic recorder.

By using data averaging techniques a response signal can be separated from noise even though visual inspection of the "raw" readouts from the EEG and ECG will show no signs of response to stimulus. Brain response to a flashing light may, for instance, be indistinguishable when viewed on an oscilloscope but will be revealed by an average response computer. So by applying this average response data to the graphic recorder in addition to the "raw" readouts from the $E E G$ and ECG, the doctor will be in a far better position to make a positive decision without delay.

## SOUND RESPONSIVE CIRCUIT

ARRANGING for electric lights to be illuminated in response to the presence of sound above a selected threshold level can be useful in many areas. The sounding of telephone or doorbells can be detected (e.g., when the user is deaf or has hi-fi headphones on); and burglar alarm systems can be devised which rely on noise detection.
In BP 1279335 Novar Electronics Corporation of Ohio, USA, also suggest that audio actuated


light can help the deaf learn to speak.

Their proposed basic circuitry is fairly simple and is shown in Fig. 1 above. An a.c. source is connected to the input terminals. The anode and cathode terminals of CSR1, being in series with the lamp and the power supply thus operate as a switch to block or pass current to the lamp.

The microphone picks up sounds, converts them into an audio output which is amplified and impressed on the gate elactrode of CSR1.

In one possible circuit the first stage amplifier transistor TR1 is coupled to TR2 with its load re sistor being the track of the 5 K control VR1. The wiper of VR1 is coupled via a $5 \mu \mathrm{~F}$ capacitor C1 to a third amplifier transistor TR3, the latter being coupled to the thyristor gate via the $32, \mu \mathrm{~F}$ capacitor.

In the absence of sound at the microphone no trigger signal is present at CSR1 gate and the lamp LP1 stays off. But any signal from the microphone is amplified and if the peaks at gate are greater than the gate firing voltage, the lamp will be illuminated during that part of each half cycle of the a.c. source in which the anode and cathode terminals of the CSR are forward biassed.
The potentiometer VR1 of course, controls sensitivity, and thus can effectively decide between either continual illumination or flickering during speech as well as basic sensitivity to telephone bells.

THE subject of this month's article is the clock generator board which has the function of generating a range of timing pulses for use throughout the arithmetic operations. A single capacitor is used to control the speed at which the whole system operates and this fact means that by running the system at slow speed, faults become relatively easy to trace.

## CLOCK BOARD

The circuit and logic of the clock pulse generating board is more interesting than. say, the shift register boards because of the diversity of the logic operations performed. The tasks which this circuit is called upon to carry out are listed below:
(a) Generate batches of ten clock pulses for the A, Z, and AD boards when instructed to do so by the programme. During addition or subtraction a single batch is required whereas during multiplication or division batches are produced continuously until detection logic decides that the arithmetic operation has been completed.
(b) Generate a single batch of n (or $10-\mathrm{n}$ ) clock pulses when instructed to do so at the beginning of the multiplication programme and the end of the dIVISION programme respectively to clock the $A, Z$, and ad boards. These clock pulses are used for decimal point positioning purposes. where n is less than ten and is determined by the setting of the decimal point thumbwheel.
(c) Provide an ungated clock pulse output to operate the programme counter and the ENTRY register normalisation system.
(d) Generate a single output pulse for every batch of ten clock pulses produced so that the number of batches can be counted during the multiplication and division programmes. This pulse to be generated during the batch generation time. to offset propagation delay problems.
(e) Generate a single output pulse at the end of each batch of clock pulses for use with the overflow detection logic.
(f) Generate an output to restart the programme sequence at the end of the series of additions or subtractions carried out during multiplication and division.
(g) Provide an isolated buffering circuit for the large load represented by the a board preset input. (This buffer is not connected with any of the other clock circuits.)

## BASIC CIRCUIT

The heart of the clock board is the circuit which produces a batch of ten pulses on demand, all other requirements being met by modification of this basic logic arrangement, which is shown in Fig. 7.1.

The "clock" shown in Fig. 7.1 is simply a square wave oscillator, and this is used to trigger a fourstage, decade counter via a single gate.
The counter will register each of the clock pulses until the tenth one sets the decade counter back to the 0000 state.

This event is detected by the output of the D counter stage falling. and is used to clock a D type flip-flop which has its data input ( $D$ ) connected to a level "1" gate output (stop clock signal). When

the counter closks the flip-flop in this way the stop -CLock input causes the flip-flop Q output to go to a "1" level and its $\bar{Q}$ output to go to a" " 0 " level.

The " 0 " level output from the $\bar{Q}$ side of the flipflop is connected back to the gate in series with the clozk input to the counter. and has the effect of preventing any further pulses from getting through to the trigger input.

If at some future time another batch of ten pulses is required. the flip-flop can be reset via its Clear input. whereupon the sequence is repeated.


Fig. 7.1. This diagram shows the heart of the clock generator board

If more than one batch of ten pulses is required the STOP CLOCK input can be left in the " 0 " state so that when the flip-flop is clocked at the end of ten pulses, its $\overline{\mathrm{Q}}$ output remains in the " 1 " state.

When sufficient batches have been produced the sTOP CLOCK input can be taken to the "l" condition so that at the end of the next batch the clock is shut off.

Note that with this system no matter how long the clock is allowed to run before being stopped the total number of clock pulses produced will always be divisible by ten, and this is obviously essential for the correct operation of Digi-Cal.

If a smaller number of clock pulses were produced during any batch, the answer produced would be in error by one or more powers of ten.

## FULL CIRCUIT

The full clock circuit is shown in Fig. 7.2 and as can be seen a considerable amount of extra gating has been added to modify the basic circuit to enable it to perform all the tasks listed earlier. To make


Fig. 7.2. The full circuit diagram of the clock generator board


Fig. 7.3. The internal logic of the SN74190 reversible counter with pin numbers of the D.I.L. package
the circuit easier to follow, the gates have not been rigidly constrained within a square box as they have been in previous circuits. That kind of approach can make the whole thing look more complicated than it really is where a lot of gating logic is employed.

## UP/DOWN COUNTER

The decade counter used on this board is not the ubiquitous SN7490 as may be expected, because although this type of counter would perform satisfactorily in the basic circuit already discussed, it could not be modified to allow it to carry out task (b).

The counter chosen is in fact the SN74190 which incorporates several extra facilities including the ability to count down as well as up, and the provision of parallel data inputs which allow the starting count to be preset to any b.c.d. number from zero to nine.

The logic diagram of this counter is shown in Fig. 7.3 and a glance at this will serve to confirm that being able to buy one of these in a 16 pin package saves a lot of time and money over having to construct such a circuit from separate gates and flip-flops!

Note that apart from the usual clock input and data outputs. there are parallel data inputs and a wide variety of control inputs and outputs whose functions are listed as follows.

## RIPPLE COUNT

 ENABLEDecides count direction (i.e. 0 to 0 or 9 to 0 )
Allows counting to take place (not required in Digi-Cal)
Presets the counter to the condition on the data inputs
Goes to a logic " 1 " if the counter contains 9 and is counting UP, or 0 and is counting DOWN
Output used for cascading SN74190 (not required in DigiCal)

## CLOCK OSCILLATOR

The clock oscillator is very simple and takes advantage of the useful properties of the SN7413 dual, four input. Schmitt trigger gate. The gates in this package have regenerative feedback applied so that they can be operated from slowly changing d.c. inputs without any spurious oscillations being produced as can sometimes occur with the standard TTL gate circuit.

The hysteresis embodied in the input characteristic of the SN7413 makes it ideal for use as a freerunning multivibrator for use between IHz and 30 MHz .

The only components required to form such a multivibrator, in addition to one of the gates in an SN7413 package, are a 330 ohm resistor and a capacitor of suitable value to set the frequency to the required range.

When Digi-Cal is completed the clock frequency will be left set to one value. but during the testing phase it is handy to be able to alter the frequency to a much lower value so that things happen more slowly.

To assist constructors in the choice of capacitor for low frequency operation. Fig. 7.4 shows a graph for reading frequency against capacitor value.


Fig. 7.4. This graph shows the relationship between the clock frequency and the value of the timing capacitor


Fig. 7.5. Detail of Fig. 7.2 showing the flip-flop and Schmitt gate used to generate a narrow pulse

## COUNTER OPERATION

Using the output of the counter to clock the $D$ type latch flip-flop is complicated in the full circuit by the need to use different clock stimuli depending on the count direction. This is taken care of by gates G10. G11, G14 and inverters N3, N4 which, depending on the state of the UP/DOWN control line salect either the D output or the Max/min output of the counter to feed the SN7474 clock input.

The operation of the counter when generating batches of ten clock pulses is the same as that of the basic circuit, but some extra logic is necessary to start the sequence.

The start clock command arrives as a long pulse from the programme circuits, and because of its length it cannot be used directly to clear BS1. If it was used for this purpose it would start the clock generation satisfactorily, but the ten clock pulses are generated so quickly that when the time came to stop the clock the START CLOCK would still be present and another batch would begin.

To overcome this problem the necessarily long start clock pulse is used to trigger a monostable circuit in the shape of BS2 and G16, which gives a very narrow output pulse which is ideal for clearing BSI.

The use of a flip-flop and a gate as a monostable seems a little strange and to make the operation of this arrangement a little clearer it has been redrawn as Fig. 7.5. The circuit relies upon the normally troublesome "propagation delays" associated with all gates and flip-flops, to allow its operation.

When the rising edge of the start clock pulse arrives it triggers the flip-flop and causes the open circuit or " 1 " condition to be clocked through from the $D$ input to the $Q$ output. The rising $Q$ output is then used to operate the flip-flop CLEAR input via the SN7413 gate which acts as an inverter and a delay element.

When the flip-flop is cleared the circuit returns to its starting condition, ready to be retriggered in the future. The output pulse width (which could be
taken from either the Q or the $\overline{\mathrm{Q}}$ output) is dependent on the delay in the gate and the delay between a clear input and a corresponding output (this varies from device to device), but gives a pulse of about 30 to 40 ns in practice.

## STOP CLOCK INPUTS

The logic decision whether or not to allow the stop CLOCK input to BSI to go to a " 1 " level is a function of five board inputs, and depends on the arithmetic operation in progress.

The logic producing the stop clock signal comprises gates G8, G9. G13, G15, and inverters N2, N5, N6. The function Code b input (which is a " 0 " level during addition and subtraction) and the nORMALISE A REGISTER input (which is a " 0 " level during the multiplication and division decimal point positioning period) have a direct effect and stop the clock at the end of one batch of pulses.

The counter equal input (which is the basic stop signal during multiplication) and the carry store negative input (which is the basic stop signal during DIVISION) are allowed to stop the clock when sufficient batches of ten pulses have been produced, and are selected according to the programme used, by the FUNCTION CODE A input. This input is a ' 0 ', level for multiply and a " 1 " level for divide.


Fig. 7.6. This diagram shows the basic A register and adder sections to illustrate how virtual left shifts are managed

## NORMALISATION

The arithmetic section of Digi-Cal does not take account of decimal fractions directly, and treats all numbers as integers. Thus, 2.5 times 2.0 is treated as 25 times 20 , giving an answer of 500 which, when displayed, appears as $50 \cdot 0$, instead of $5 \cdot 0$ as required.

The answer reached in this way obviously needs correction, and this is achieved by shifting it to the right by the number of decimal places selected on the thumbwheel. This part of the programme sequence is called normalisation, and is achieved in a different way during division.

Division of 2.5 by 0.5 leads to an initial answer of $0 \cdot 5$, which suggests a left shift of one place is required, but in practice it is necessary to left shift not the ANSWER, but the DIVIDEND, before the arithmetic process starts. In this way, 2.5 divided by 0.5 becomes $25 \cdot 0$ divided by $0 \cdot 5$, which gives the correct answer of 5.0 without correction.

Left and right shifts such as those described above obviously require a batch of less than ten clock pulses, and this is where the preset inputs of the SN74190 are used to advantage. Before looking at

## CLOCK GENERATOR BOARD



Fig. 7.7. Layout of the components on the clock generator board and function of the edge contacts. Also shown is the wiring details of the edge connector (below left)

## COMPONENTS . . .



| CLOCK GENERATOR BOARD |
| :--- |
| Resistors |
| R71 $330 \Omega \frac{1}{4} \mathrm{~W} \pm 10 \%$ carbon |
| Capacitors |
| C30 $10 \mu \mathrm{~F} 15 \mathrm{~V}$ elect. |
| C31 $0.047 \mu \mathrm{~F}$ |
| C32 see text |
| Integrated Circuits |
| IC91 $\quad$ SN7413 |
| IC92 $\quad$ SN7400 |
| IC93 $\quad$ SN7474 |
| IC94 $\quad$ SN7404 |
| IC95 |
| IC96 SN7400 |
| IC97 |
| IC98,IC99 |

## Printed Circuit Board

Type DL109/22 (Shirehall)
the generation of these shortened batches, it is necessary to look at just how the normally rightshifting a register is made to shift left.

## VIRTUAL LEFT SHIFTS

The way in which left shifts are achieved is best studied with the aid of Fig. 7.6 which shows the basic Digi-Cal a REGISTER and adDER.

The requirement is to shift the number 1234.56 two places to the left, because division is in progress and the thumbwheel is set to two.

To turn the a register into a bi-directional type at the hardware level would require considerable numbers of extra gates, not to mention the preclusion of the use of SN7496 devices, and for this reason any such move is undesirable.

Fortunately by employing what amounts to a programming "dodge" it is possible to turn a hardware right shift into a virtual left shift, and this is achieved by clocking the register with 10 minus $n$ pulses.

By applying only eight pulses to the register in Fig. 7.6, and by leaving the output of the ADDER unmodified by the $z$ REGISTER input, the $A$ data recirculates until it ends up as $123456 \cdot 00$.

This system relies on the fact that there are at least $n$ zeros to the left of the number to be shifted, and provides one reason why the $A$ and $z$ registers are ten digits long.

## GENERATING LESS THAN TEN CLOCK PULSES

To generate the $n$ pulses needed during the multiplication normalisation the counter preset COUNT input is taken to a " 0 " level which loads the true thumbwheel code into the SN74190, e.g. DCBA equals 0011 for a setting of three.

Next, the COUNT DIRECTION input is taken to a "1" level to set the count direction to DOWN, and the Start clock input initiates the sequence.

The counter will then count $2,1,0$, whereupon the max/min output of the SN 74190 will go to " 1 " clocking BS1 and stopping the clock via G2, only three pulses having been produced in all.

To generate the 10 minus $n$ pulses required during division normalisation the counter is loaded with the same code as before but in this case the count direction is set to UP, and the sequence for the example of three is, $4,5,6,7,8,9,0$, producing a total of seven pulses. The $D$ output of the counter is used to trigger BS! in this case, not the max/min output.

## BOARD OUTPUTS

There are several outputs from the clock board, some of which (such as the $A$ and $z$ register buffered clock lines), are self-explanatory. Gate G7 is used to detect counts six and seven to give a negativegoing edge after clock seven to trigger the counter boards used during multiplication and division. A trigger for this purpose is produced early in the sequence to allow the COUNTER boards and COMparator to register an equality during multiplicaTION and still have time to stop the clock at the end of its run.
This output is called plus/minus complete. Gate G12 is used to buffer the pulse produced after each
batch of ten clock pulses to give the END OF TEN output required by the error detection, or OVERFLOW logic.

The restart output is taken from the Q output of BSl and is used to restart the programme during multiplication or division when the series of additions or subtractions have been completed.

The a register preset buffer gates are included on this board for convenience, and have no logic connection with it. They are parallel connected to provide the high drive capability required by the PRESET function of the SN7496.

## CONSTRUCTION

This circuit is built on a DLI09/22 card, which has a 22 -way edge contact on the printed side of the board. The component layout and edge contact wiring is shown in Fig. 7.7.

Wiring-up follows the same pattern as was outlined for previous boards.

The value of the capacitor used to set the clock frequency is about $3,000 \mathrm{pF}$ for the completed DigiCal (three $001 \mu \mathrm{~F}$ capacitors in series were used in the prototype) but during testing procedure it is often convenient to use a very slow clock to enable the exact operation of the programme for instance to be traced using only the display and perhaps a voltmeter as indication.

The prototype has been tested with values as high as $1,000 \mu \mathrm{~F}$ and still operated correctly, if rather slowly! For this reason it is better not to wire in a capacitor permanently at this stage since a suitable component can be connected when required.

## Next month: The Adder Board

## Pollits nilishl

## I.C. INTERCOM (October 1972)

It has been brought to our notice that the PA234 i.c. is no longer available. A suitable alternative (not substitute) is the LM380, a 2 W amplifier with its gain preset at 50 . It requires fewer external components than the PA234 (R4, R5, R6, and C5 are not necessary) as shown below.

Minor alterations to the Veroboard layout are required. Pins 3.4. 5, 10, 11 and 12 must be connected to a heatsink for 2 W . It is recommended that a piece of copper be soldered to these pins under the panel if the i.c. is to be used at high power for any length of time. (This i.c. will drive an 8 ohm speaker.)

The LM380 is available ( $\mathfrak{f 1} \cdot 45$ ) from D.T.V. Group Ltd., 126 Hamilton Road, London, S.E.27.


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## QUALITY COMPONENTS - BUDGET PRICES



A look at some of the hi fio developments seen and heard at Olympia By M. A. COLWELL



The horn loaded loudspeakers from Lecson Audio are available in a variety of finishes

ARE WE so immersed in the technicalities of hi fi equipment that we are sometimes guilty of spurning the real purpose of that equipment?

This poignant remark heard at this year's Audio Fair at Olympia is perhaps more significant in its implications than in the statement itself. It was a further cause for ponderance on tasting some of the lectures held simultaneously with the exhibition. These were enthusiastically attended and the presentation by the speakers was generally of a high standard. But one must hold some sympathies with the lay public who in some cases found the technical content some justification for not staying to the end.

On further analysis one can conclude that those keenly interested in hi fi equipment are extremely knowledgeable on the technical aspects. So much so that this is largely responsible for the cool reactions so far to quadraphonic reproduction in this country.
There has so far been a continuous air of "wait and see how it goes," while the Japanese persistently try to convince the British public that their systems are "proof of the pudding." If one was to assess quadraphonic sound on what one hears at Olympia then there can be little surprise at disillusionment, with comments like . . "But who wants to sit in the middle of the orchestra."

## SIMULATED AMBIENCE

To sell quadraphonic sound, not only does the cost have to be fully justified, but also the ultimate accommodation in the home has to be compatible with normal domestic requirements. With necessary furnishings to blend with, both acoustically and visually, successful four-channel ambience is largely dependent on room dimensions and acoustic properties. A pseudo mock-up in a crowded stereotype
cubicle does not give a fair indication to the listener of what quadraphonic ambience should be like in a domestic setting and all too often these demonstrations were valueless. As one exhibitor put it: The speaker system is only as good as the room that loads it. The addition of ten or twenty listeners in the room adds further influence.

Only a trickling: of British companies are offering four-channel facilities, usually in the form of a matrix decoder fed from two-channel equipment. The merits of this method lie only in the convenience of adaptability from existing equipment to suit "quadraphonic" recordings. The major drawback is some loss of channel separation. True quadraphonic ambience is only really authentic when derived from totally independent channels. The CD-4 discrete recorded four-channel system is the recommended version that uses this method.

Those offering quadraphonic equipment at the Audio Fair include the following (the code letters indicate equipment origin): J-Japan. WG-West Germany, UK-United Kingdom, USA-United States of America, DM-Denmark.

## BRITISH FOUR-CHANNEL EQUIPMENT

Goodmans Module 90 tuner amplifier (UK)-an impressive piece of equipment with excellent styling. This unit receives m.w. and f.m. bands by preset or variable tuning and contains a $45+45$ watt r.m.s. amplifier with less than $0 \cdot 1$ per cent total harmonic distortion.
The model shown in our photograph is the white finish version and has elegant push-button arrays, twin level meters and slider controls. The "4D ambiophonic capability" will require two additional speaker units, while the main circuitry incorporates many of the features of the slightly
more expensive Models One-Ten and Module 80 (f.m.). Four f.m. band presetting controls are provided at the bottom right-hand end which are selected by four push-buttons above them. Two power output meters are also fitted.

Compared with some units this represents good value at $£ 121$ (white finish).

## FOREIGN QUAD

Grundig (WG) have an integrated tuner/amplifier complete for four-channel sound and including all the necessary decoder and amplifier circuitry ready for connection to four speakers. This is called the "Studio 2000b," has all the features expected of its kind, but is priced at $£ 238.60$, which includes a Dual 1216 transcription auto-changer fitted with a Shure M75 MBD magnetic cartridge.

The tuner has seven v.h.f. pre-selectors, each with its own tuning scale and short, medium and long wave reception. This might be classed as budget equipment of superior quality although the distortion figures ought to be better.

An alternative version without the transcription unit and of lower output ( 12.5 W r.m.s.) is the Grundig RTV 800, and the RTV 900 at 25 W r.m.s. per channel is similar in appearance. Similar push buttons are used as in the Goodmans unit and slider controls are also preferred. The photograph shows the RTV 800 with the individual tuning scales to the left of the main scale.

All three of these models are large, at least 22in long. the Model 2000b being larger. The RTV 800 is priced at $£ 135 \cdot 80$ and the RTV 900 at $£ 177.25$.

Other ranges of quadraphonic equipment shown were from National Panasonic (J), JVC Nivico (J), Pioneer (J), Sansui (J), Sanyo (J), Siemens (WG), Sony (J), most of whom have demonstrated this technique before in some form. Golding Skandia (J) also had cartridge players quoted as "simulated quadphonic" at various prices.


Marantz Model 19 stereo receiver


The layers of the isodynamic headphone structure by Rank Wharfedale. Note the printed circuit style diaphragm at D

## ISODYNAMIC HYBRID

In taking an overall view of the equipment on show it is evident that manufacturers are coming close to the ultimate performance versus cost factor and innovations such as have been regular occurrences in the past are now very thin on the ground.

Probably one of the most significant commercial developments though shown this year is the Isodynamic Stereo Headphone made by Rank Wharfedale. High quality stereo headphones have so far been somewhat heavy for long term listening, but this one weighs only 13 ounces ( 375 grammes). It is a semi-hybrid whereby the whole diaphragm is driven evenly and freely as in the electrostatic principle, but is influenced by an anisotropic ceramic/ synthetic rubberised sheet magnetised perpendicular to the diaphragm.

The diaphragm is a polymide film with flexible printed circuit style metallic grid. The electromagnetic influences caused by the sound signal passing through this grid, in conjunction with the magnetised rubber, make the diaphragm move.

Subjective measurements on these phones show no violent variation in spectral response and the frequency response using a standard "artificial ear" shows -3 dB points at 40 Hz and 15 kHz . The price is competitive at $£ 19 \cdot 95$.

## RAPID DEVELOPMENT

With the expansion of stereo broadcasting to Radio 2 (from early November) and the likelihood of commercial sound broadcasting to come, one would have expected to see a strong emphasis on advanced tuner developments using integrated circuits. There were surprisingly few; the general trend seems to be to integrate the complete tuner with the amplifier in one case. Very few separate tuners were evident.
However, Lecson Audio (a promising newcomer) have in less than 12 months produced some superb high quality equipment in revolutionary new styling. The tuner TF1 is a self-powered unit with preset v.h.f. tuning and variable tuning controls for stereo or mono broadcasts.
The controls take the form of sliding bars matching the case styling. The performance is specified as $1 \cdot 5 \mathrm{NV}$ sensitivity for 30 dB quieting; a.m. rejection 42 dB ; signal-to-noise ratio better than 60 dB ; audio frequency response $\pm 1 \mathrm{~dB} 20 \mathrm{~Hz}$ to $14 \cdot 5 \mathrm{kHz}$.

## LOW DISTORTION

The matching control unit pre-amplifier, also selfpowered with slide controls and electronic switching. is a two-channel stereo unit with two four-channel configurations for a supplementary matrix decoding surround-sound unit, to be made available. The performance is claimed to be so impressive that the sceptics will probably need convincing.
There are the usual input facilities to be expected with signal-to-noise ratios of 70 dB (magnetic pickups) and 80 dB for others. Outputs supply 500 mV into $1 \mathrm{k} \Omega 2$ loads for pick-up and radio; 100 mV for tape; 2W into 8 ohms for headphone use. Distortion is quoted as typically less than 0.02 per cent; frequency response $\pm 0.5 \mathrm{~dB} 10 \mathrm{~Hz}$ to 25 kHz .

The power amplifiers AP1 and AP2 boast distortion figures of less than $0.005 \%$ at 1 kHz at 35 W , less than $0.02 \%$ at 50 W and better than $0.05 \%$

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## TRANSFORMERS

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LT3 $12 \mathrm{~V}, 1-5 \mathrm{~A}-87 \mathrm{p}$ plus 26 p p. \& LT4 12V. 3A-21.32 ptus 30p LT5 9-0-9V, 0.5A-75p plus $21 p$ LT6 12-0-12V. IA-95p plus 26p LT7 30-0-30V, IA- 11.87 plus 30 p . Multi-tapped
MT30/2 0-12-15-20-24-30V, 2A$\begin{array}{ll}\text { MT60/I } & 0-5-20-30-40-60 \mathrm{~V} \text { p. IA } \mathrm{F} \text {. }\end{array}$
 Charger
 CT/03 4A-f1. 60 plus 30p p.
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SEMICONDUCTORS, etc.
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 50 p
Bridge rectifier-200 P.I.V., $2 \cdot 0 \mathrm{~A}$,
50 p Transistor sockets-7p
D.1.L. I.C. sockets-14 pin, 20p: 16 pin, 20 p
IN $4001-50$ P.I.V.. $1 \cdot 0 A, 6 p$,
IN 4002 - 100 P.I.V., $1 \cdot 0 A, 7 p$ IN 4002 - 100 P.I.V., I OA, 7p
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 $\begin{array}{lll}\text { GBI3 } 6 i n & \text { in } 2 i n & 52 p \\ \text { GBI4 } & \text { in } \\ \text { in }\end{array}$
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EQUIPMENT CASES in plain aluminium with sloping front panel.
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mer finished, 25p extra.

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n plain aluminium, ideal for mixers,
Type W. A B C D Price p. \& p.
 $\begin{array}{lllllll}G 821 & 10 & 9 & 3 \frac{1}{2} & 3 & 61.58 & 30 p \\ \text { GB22 } & 12 & 9 & 3 \frac{1}{2} 2 & 3 & 61.72 & 30 p\end{array}$
 |d |
b $\square$

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for microphones or guitars. Mounted on printed circuit panel and complete with t inch jack socket. Requires Complete with connection data. com
65 p .

ELECTROLYTICS

| - |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $1 \mu \mathrm{~F}$ | 450 V | 19p | 1,000 F | 25 V | 27 |
| $2 \mu \mathrm{~F}$ | 450 V | 20p | 1,000 $\mathrm{\mu F}$ | 50 V | 42 p |
| $4 \mu \mathrm{~F}$ | 350 V | 14p | 2,000 FF | $25 V$ | 39 |
| $8 \mu \mathrm{~F}$ | 450 V | 17p | 2,000 FF | 50 V | 53 |
| $16 \mu \mathrm{~F}$ | 450 V | 18p | 2,500 FF | 25 V | 45 |
| $25 \mu \mathrm{~F}$ | 25 V | 7p | 2,500 F | 50 V | 60 |
| $25 \mu \mathrm{~F}$ | 50 V | 10p | $3,000 \mu \mathrm{~F}$ | 25 V | 48 |
| $32 \mu \mathrm{~F}$ | 450 V | 27p | $5.000 \mu \mathrm{~F}$ | 25 V | 60 p |
| $50 \mu \mathrm{~F}$ | 50 V | 10p | 5,000 F | 50 V | ¢ 1.10 |
| $100 \mu \mathrm{~F}$ | 25 V | 10p | $8-8 \mu \mathrm{~F}$ | 450 V | 18 |
| $100 \mu \mathrm{~F}$ | 50 V | $11 p$ | $8-16 \mu \mathrm{~F}$ | 450 V | 20 |
| $250 \mu \mathrm{~F}$ | 25 V | 14p | $16-16 \mu \mathrm{~F}$ | 450 V | 27 |
| 2SOHF | 50 V | 17p | $16-32 \mu \mathrm{~F}$ | 450 V | 63 |
| 500 HF | 25 V | 18p | 32-32 $\mu \mathrm{F}$ | 450 V | 49p |
| $500 \mu \mathrm{~F}$ | 50 V | 25p | 50-50 F F | 350 V | 38p |

MINIATURE ELECTROLYTICS

| $1 \mu \mathrm{~F}$ | 63 V | 6p | 47 $\mu \mathrm{F}$ | 16 V | 7p |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $2 \cdot 2 \mu \mathrm{~F}$ | 63 V | 6p | 47, F | 25 V | 6 p |
| $3 \cdot 3 \mu \mathrm{~F}$ | 63 V | ${ }^{6 p}$ | $68 \mu \mathrm{~F}$ | 16 V | ${ }^{6 p}$ |
| $4.7 \mu \mathrm{~F}$ | 63 V | $6 p$ | $100 \mu \mathrm{~F}$ | 10 V | 6p |
| $8 \mu \mathrm{~F}$ | 40 V | 7p | $220 \mu \mathrm{~F}$ | 16 V | 7 p |
| $10 \mu \mathrm{~F}$ | $25 V$ | 6p | $330 \mu \mathrm{~F}$ | 16 V | $11 p$ |
| $10 \mu \mathrm{~F}$ | 64 V | 7p | $470 \mu \mathrm{~F}$ | 10V | 119 |
| $16 \mu \mathrm{~F}$ | 40 V | 7p | 1,000 F | 16 V | 19p |
| $33 \mu \mathrm{~F}$ | 16 V | 6p | 1,500 $\mu \mathrm{F}$ | 16 V | 23 |

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$5 \mathrm{k} \Omega, 10 \mathrm{k} \Omega, 25 \mathrm{k} \Omega, 50 \mathrm{k} \Omega, 100 \mathrm{k} \Omega, 250 \mathrm{k} \Omega$
SLIDER CONTROLS, 87 mm .
Single, 44 p ; Tandem, $55 \mathrm{p} .10 \mathrm{k} \Omega, 25 \mathrm{k} \Omega$. $50 \mathrm{k} \Omega, 100 \mathrm{k} \Omega$, log. o

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All $5 \%$, high-stability, E12 values. $\frac{1}{4} W$, Ip; AW, itpigh-st 4D; 2W, 6p
Wire.wound
5W. 10p: 10W, 12p

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24 way, 0.15 (32in)-34p

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| D.IN.N. 2 pin (speaker) | 10p |
| D.I.N. 3 pin | 13p |
| D.I.N. 4 pin | 14p |
| D.I.N. 5 pin, $180^{\circ}$ | 13p |
| D.I.N. 5 pin, $240^{\circ}$ | 15p |
| D.I.N. 6 pin | 15p |
| jack, $2 \frac{1}{2} \mathrm{~mm}$ unscreened | 9 p |
| Jack, $2 \frac{1}{2} \mathrm{~mm}$ screened | 10p |
| Jack, $3 \frac{1}{2} \mathrm{~mm}$ unscreened | 8p |
| Jack, $3 \frac{1}{2} \mathrm{~mm}$ screened | 12p |
| Jack, tin unscreened | 12p |
| Jack, tin screened | 20p |
| Jack, stereo, unscreened | 20p |
| Jack, stereo, screened | 35p |
| Phono, plastic top | 5p |
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| Wander, red or black | 3 p |
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| Car aerial | 14p |
| Co-axial | 17p |
| D.I.N. 2 pin (speaker) | 15p |
| D.I.N. 3 pin | 16p |
| D.I.N. 5 pin, $180^{\circ}$ | 16p |
| D.I.N. 5 pin, $240^{\circ}$ | 16p |
| Jack, $3 \frac{1}{\text { mmm }}$ | 15 p |
| Jack, $\frac{1}{4}$ in screened | 49 p |
| Jack, stereo, screened | 34 p |
| Phono, plated metal | 14p |

$14 p$


Car aerial Co-axial, flush D.I.N. 2 pin (speaker) D.I.N. 3 pin 180 240 ack, $2 \frac{1}{2} \mathrm{~mm}$
Jack, $3 \frac{1}{2} \mathrm{~mm}$
Jack, tin unswitched Jack, tin unswitched
Jack, tin switched Jack, stereo, switched Phono, single Phono, 2 on a strip
Phono, 3 on a strip Phono, 3 on a strip
Phono, 4 on a strip Wander, single, red or black 10 p Wander, twinstrip 7 p Banana 4 mm red, or black $6 p$

## CAPACITORS

## $\begin{array}{ll}2 \cdot 2 \mathrm{pF} & 500 \mathrm{~V} \\ \text { 3.3pF } & \mathrm{S} / \mathrm{M} \\ & 500 \mathrm{~V} \\ \mathrm{~S} / \mathrm{M}\end{array}$



| F F | 500 V |  | 15p |
| :---: | :---: | :---: | :---: |
| $0.003 \mu \mathrm{~F}$ | 500 V | Cer. | p |
| $0.0033 \mu \mathrm{~F}$ | 125 V | P.S. | p |
| $0.0033 \mu \mathrm{~F}$ | 500 V | Poly. | $8 p$ |
| $0.0033 \mu \mathrm{~F}$ | 1,000V | MOC | $6 p$ |
| $0.0036 \mu \mathrm{~F}$ | 500 V | S/M | 5p |
| $0.0047 \mu \mathrm{~F}$ | 125 V | P.S. | 9 p |
| $0.0047 \mu \mathrm{~F}$ | 500 V | Poly | ¢p |
| $0.0047 \mu \mathrm{~F}$ | 500 V | S/M | 20p |
| $0.0047 \mu \mathrm{~F}$ | 1,000V | MDC | $6 p$ |
| $0.005 \mu \mathrm{~F}$ | 100 V | Mylar | p |
| $0.005 \mu \mathrm{~F}$ | 500 V | Cer | pp |
| $0 \cdot 0068 / 2 \mathrm{~F}$ | 125 | $P$. | 101p |
| $0.0068 \mu \mathrm{~F}$ | 500 V | S/M | 30p |
| $0 \cdot 0068 \mu \mathrm{~F}$ | 500 V | Poly. | 6p |
| $0.0082 \mu \mathrm{~F}$ | 125 V | P.S. | $101 p$ |
| $0.0082 \mu \mathrm{~F}$ | 500 V | S/M | 30p |
| $0.01 \mu \mathrm{~F}$ | 18 V | Dise | 4p |
| $0.01 \mu \mathrm{~F}$ | 125 V | P.S. | 101 P |
| $0.01 \mu \mathrm{~F}$ | 160 V | Poly. | p |
| 0.0112 F | 250 V | M.F. | 3 p |
| $0.01 \mu \mathrm{~F}$ | 400V | Poly | 3 p |
| $0.01 \mu \mathrm{~F}$ | 500 V | Ce | p |
| $0.01 \mu \mathrm{~F}$ | 500 V | S/M | 30p |
| $0.01 \mu \mathrm{~F}$ | 600 V | MDC | 7 p |
| $0.01 \mu \mathrm{~F}$ | 1,000V | MDC | 9p |
| $0.015 \mu \mathrm{~F}$ | 160 V | Poly. | 3p |
| $0.015 \mu \mathrm{~F}$ | 400 V | Poly. | 3 p |
| $0.02 \mu \mathrm{~F}$ | 100 V | Mylar | ${ }^{3} \mathrm{P}$ |
| $0.022 \mu \mathrm{~F}$ | 18 V | Disc | 5p |
| $0.022 \mu \mathrm{~F}$ | 250 V | M.F. | p |
| $0.022 \mu \mathrm{~F}$ | 400 V | Poly. | 3 p |
| $0.022 \mu \mathrm{~F}$ | 600 V | MDC | 1 P |
| $0.022 \mu \mathrm{~F}$ | 1.000 V | MDC | 10p |
| 033 $\mu \mathrm{F}$ | 250 V | M.F. | 4p |
| $0.033 \mu \mathrm{~F}$ | 400 V | Poly, | 4p |
| $0.047 \mu \mathrm{~F}$ | 2 V | Disc | $6 p$ |
| $0.047 \mu \mathrm{~F}$ | 160 V | Poly. | ${ }^{3} \mathrm{p}$ |
| $0.047 \mu \mathrm{~F}$ | 250 V | M.F. | 3p |
| $0.047 \mu \mathrm{~F}$ | 400 V | Poly. | 4p |
| $0.047 \mu \mathrm{~F}$ | 600 V | MDC | 8 p |
| $0.047 \mu$ | 1.000 V | MDC | 10p |
| $0.1 \mu \mathrm{~F}$ | 30 V | Dise | 6p |
| $0.1 \mu \mathrm{~F}$ | 250 V | M.F. | 4 p |
| $0 \cdot 1 \mu \mathrm{~F}$ | 400 V | Poly. | p |
| $0.1 \mu \mathrm{~F}$ | 600 V | MDC | 10p |
| $0.1 \mu \mathrm{~F}$ | 1.000 V | MDC | 14p |
| $0.15 \mu \mathrm{~F}$ | 250 V | M.F. | 5p |
| $0.22 \mu \mathrm{~F}$ | 160 V | Poly. | $6 p$ |
| $0.22 \mu \mathrm{~F}$ | 250 V | M.F. | 5p |
| $0.22 \mu \mathrm{~F}$ | 400 V | Foil | $10 p$ |
| $0.22 \mu \mathrm{~F}$ | 1.000 V | MDC | 15p |
| $0.33 \mu \mathrm{~F}$ | 250 V | M.F. | \% |
| $0.47 \mu \mathrm{~F}$ | 250 V | M.F. | 8p |
| $0 \cdot 47 \mu \mathrm{~F}$ | 400V | Foil | 15p |
| 0.47 $\mu \mathrm{F}$ | 1.000 V | MDC | 25p |
| $1.0 \mu \mathrm{~F}$ | 250 V |  | 15p |
| Note: <br> $S / M=$ silver mica $1 \%$ tol. P.S. = polystyrene $2 \frac{1}{2} \%$ tol. MDC -a.c. rating $=300 \mathrm{~V}$. M.F. $=$ Mullard min. foil. Cer. =ceramic. |  |  |  |
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(60/40 alloy)
22swg. 22p

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Model 3A. Strips insulation from cable or flex without nicking wire. 4 different settings, 4\&6BAspanner ends, ground cutting edges Price 32p. Also available de luxe Model 8.

Price 58p.
From Electrical and Hardware Shops. If unobtainable, write to: Multicore Solders Ltd., Hemel Hempstead, Herts.
at all other frequencies and power levels. Transient distortion effects are claimed to be non-existent, while noise is quoted at less than 90 dB below 35 W .

The two models APl (50W per channel) and AP2 (minimum 70W per channel) are housed in decorative cylindrical cases. Prices match performance: The tuner is $£ 70, \mathrm{ACI}$ control unit $£ 95$, AP1 $£ 75$, AP2 £90, and the excellent horn loudspeaker introduced earlier this year $£ 89$. Distribution is through appointed dealer only network.

The new AR1214 a.m./f.m. stereo receiver from Heath (Gloucester) Ltd. has an integral $15 \mathrm{~W}+15 \mathrm{~W}$ r.m.s. power amplifier and. as in the case of the P.E. Gemini Tuner, uses two integrated circuits, two ceramic filters, a phase locked multiplex demodulator, and a pre-assembled ready aligned f.m. tuning head to provide $2,4 \mathrm{~V}$ sensitivity. The kit price is $£ 69 \cdot 80$.

## PRICED PERFORMANCE

Prices of equipment generally at the Fair were high when compared with the forerunners; some may be justified, others perhaps taking advantage of a booming market. It is a strange fact that the high quality goods attract some very high prices.

It is now not unusual to find a single unit costing more than $£ 100$ and some tuner/amplifiers combined at more than $£ 200$. One even goes as high as $£ 775$ and, when $£ 125$ is donated to Government funds in the form of purchase tax, the buyer will be asked to find $£ 880$-almost the price of a reasonably comfortable family car.

This particular model is the Marantz Model 19 f.m. stereo receiver. The Marantz demonstrator went to great pains to illustrate that all their amplifiers gave an identical sound (and so it seemed to be at Olympia); the reason for the variation in prices was said to be in the facilities offered.

Comparing this Model 19 with their a.m./f.m. stereo receiver Model 2215 at $£ 186$, an immediate difference is apparent: the 2215 has an f.m. sensitivity of 4.5 mV against 1.8 mV , stereo separation at 30 dB against 45 dB , t.h.d. and intermodulation distortion at $0.5 \%$ against $0.15 \%$.

There are others, but do they really add up to nearly $£ 600$ worth of extra materials or technical expertise? They also seem to believe that users will frequently need to check amplifier performances on the built-in oscilloscope in the Model 19. Surely this should not be necessary at this price. This American company builds its equipment in Belgium and distributes in the UK through Pyser-Britex (Swift) Ltd., of Edenbridge, Kent.

By comparison with some British units the cost/ performance factor of these units and the new Model 1030 and 1060 amplifiers is generally on the high side.

## SERVO TRANSCRIPTION

One would not have thought that there was much more room for development on the transcription unit, but $\mathrm{B} \& \mathrm{O}$ attracted considerable attention with their Model 4000 , mainly due to the electronically controlled tangential arm.

Disc cutting heads traverse the disc in a straight line along the radius by using the "lathe screw" principle of motion. The 4000 pick-up does likewise, but instead of having only one arm, it has two, one carrying a photocell which senses the presence, size, and specified playing speed of the disc.
 arm transcription unit with servo control by Bang \& Olufsen

Pye 9145 stereo cassette recorder with DNL


In conjunction with a servo motor the pick-up lathe screw is driven in synchronism with the rotation of the disc, so avoiding drag friction and skating problems. Preset controls on a panel are set to feed basic data to enable correct operation to be carried out. Any error in setting up will inhibit operation.

Provided specially for this pick-up arm, the new SP15 cartridge has a sealed-in stylus which is factory fitted and tested as being the perfect match for the arm supplied. The price for the complete transcription unit is $£ 179.50$ (teak), $£ 180 \cdot 50$ (rosewood), and replacement cartridges SP15 are $£ 36 \cdot 08$.

## IN BRIEF

Following the success of the DMI Monitor loudspeaker system by B \& W Electronics at $£ 37 \cdot 39$, they have now introduced the DM4 which has greater power handling capacity and extended bass response down to 30 Hz . This model sells at $£ 49 \cdot 66$.

BSR have produced an eight-track stereo tape cartridge player, the TD8-3V at $£ 27.99$.

New plinth systems have been designed and produced by SME, the Model 2000, which is available in component form to suit various Thorens transcription units and SME pick-ups. Prices vary, but for a complete plinth with Perspex cover you can expect to pay in the region of $£ 50$.

Garrard (part of the Plessey organisation) have also produced plinths with Perspex lids. These "modules" are supplied complete with the transcription units for which they are made, prices ranging from $£ 23 \cdot 49$ for the SP25 Mk 3 (less cartridge) to $£ 59.07$ for the Z100S (less cartridge), including purchase tax.

Philips introduced their N2510 high quality stereo cassette deck recorder for chromium dioxide tape. It has twin meters, and sleek slider and push button controls.

The well-known Philips dynamic noise limiter circuit is included in the Philips N2506 and the equivalent Pye model 9145 stereo cassette recorders costing $£ 65$.

Other less expensive items of equipment were on show, but it seems that, like most things, new products are an accepted method of attracting higher prices, but that is another subject-not for this page.


TRANSISTOR AUDIO AND RADIO CIRCUITS Published by Mullard
283 pages, $8 \frac{1}{2}$ in $\times 5 \frac{1}{4} \mathrm{in}$. Price $£ 1.80$

The great popularity of the first edition of this excellent book published in 1969 has led Mullard to update and augment the information, and present it in a second edition. The appeal of the first edition must be attributed to the wealth of useful and really practical circuits presented. All the circuits are of proven design and the performance figures quoted are taken from the prototypes rather than merely theoretical calculations.

In addition to the established designs for 10 W and 25 W amplifiers there are three new circuits for 15 W , 35 W , and 50 W amplifiers together with a preamplifier for these circuits.

Other additions to the first edition are a radio receiver, and amplifiers using integrated circuits. There is also a new chapter on loudspeakers, which discusses choosing a suitable speaker and design features of enclosures for high quality systems.

Other sections describe tape recorders with a design for a 4W machine, a circuit for a high quality f.m. tuner, and a chapter on a.f. amplifiers for car radios. Full details of all components are given as well as suppliers for the more difficult to obtain devices. No constructional details are given although where special care needs to be taken, as in the case of stereo amplifiers, this is more than adequately discussed.

Though not designed for the beginner this is a book from which everyone can learn, and a price that anyone can afford makes this an excellent, practical, reference work.
S.R.L.

HI FI YEAR BOOK-1973

## Edited by C. Sproxton <br> Published by IPC Electrical-Electronic Year Book Ltd.

464 pages, $8 \frac{3}{4}$ in $\times 6$ in. Price $£ 1.50$

AFURTHER 88 pages on top of last year's volume, packed with many more items of equipment and accessories for the audiophile. For those new to this regular work, the Hi Fi Year Book catalogues technical specifications and photographs of a wide range of domestic and studio hi fi equipment, following four introductory articles on the state of the art. The articles this year include an insight into specifications jargon, a selection of pre-recorded cassette tapes, an up to date look into quadraphony -its techniques and purpose, and a survey of loudspeaker development.

Included in the classified directory are added quadraphonic equipment, tape transport machines, decks, and players, tables on tape playing times and v.h.f. transmitters. The standard of production is well maintained, although equipment news released after September 1972 may not be included (for some of these refer to our Audio Fair report in this issue).

## BEGINNERS GUIDE TO TELEVISION

By Gordon J. King
Published by Newnes-Butterworths
211 pages, 5 in $\times 7$ 옹in. Price $\mathbf{£ 1 . 6 0}$.

DOMESTIC television receivers have for long had a special kind of fascinating appeal to electronics followers, probably due to the various "tricks" in achieving a satisfactory performance with some pruned down economic circuitry. Since transistors have to a substantial extent replaced valve circuits. I was surprised to find little more than a page devoted to this aspect, but then the majority of this book is concerned with general principles rather than detailed circuit theory. In the examples of valve circuits described, only the basic timebase and pulse circuits are explained, the remainder of the excellent diagrams are mainly illustrative of methods of system operation. The photographs are also of excellent quality, although one or two would probably rate as superfluous.

It is an interesting fact that the 32 -page chapter on colour television forms a useful introduction to the colour techniques used, but in the U.K. and Europe the N.T.S.C. system is really only of academic interest. If this system is worth inclusion then, for comparative discussion, so too is SECAM. PAL is the main domestic system, and is briefly described, including delay line techniques: greater detail would have enhanced the value of this book to the service engineer.

Whilst being somewhat critical of contents selection, one must compensate by complimenting the down to earth writing style and presentation of this popular author. This book is still to be recommended on this account if it suits your requirements, but beginners beware-you do need some prior knowledge of general radio and electronics theory for some parts.
M.A.C.

## READER'S DIGEST REPAIR MANUAL Published by The Reader's Digest Association 704 pages, $7 \frac{1}{4}$ in $\times 10 \frac{1}{2}$ in. Price $£ 6.30$

THOUGH the price of this book may be somewhat discouraging, the amount of information contained within its pages is really enormous, covering everything from jewellery to caravans. All practical details are explained with the aid of a multitude cf clear, "comic-strip" type drawings and the section on renovating and restoring contains some excellent colour photographs.

To elucidate further on the contents, the main areas covered are: house repairs and decoration; renovating and restoring; repairs around the home and in the garden; electrical equipment; in the garage: and a reference section giving miscellaneous information and advice.

The section on electrical repairs assumes no technical knowledge and thus only deals with "mechanical" work. The chapter on cars contains some useful information on fitting suppressors to improve radio reception.

This book is not only for the do-it-yourself man but also includes lots of information that women would find very useful, not to mention the children.
S.R.L.

## THIS IS THE FIRST PAGE OF THE GREAT BI-PAK SECTION

## BRAND NEW FULLY GUARANTEED DEVICES

reference

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## Sinclair Project 60

## Stereo 60


pre-amplifier/control unit

The versatility of Project 60 high-fidelity modules is well demonstrated in this excellent unit. It provides the facilities essential to good stereo and will enhance the quality of any system it is used with, whether Project 60 or any any othertop line power amplifiers. Compact, yet robustly constructed, the unit is easily panel mounted and will operate satisfactorily from 18 to 35 volts supply. Silicon epitaxial transistors are used throughout to achieve a very high signal to noise ratio with excellent separation between channels. Distortion at maximum output is barely $0.02 \%$ with magnetic p.u. input. Accurate equalisation is provided for all inputs, which are selected by push buttons. For maximum effectiveness, the Sinclair A.F.U. is recommended for use with the Stereo 60 pre-amp/control unit. A comprehensive manual supplied with Project 60 modules makes installing and connecting easy and ensures best possible results from your system.

## Super IC. 12 <br> Integrated circuit <br> high fidelity amplifier



Having introduced Integrated Circuits to hi-fi constructors with the IC.10. the first time an IC had ever been made available for such purposes, we have followed it with an even more efficient version, the Super lC. 12 , a most exciting advance over our original unit. This needs very few external resistors and capacitors to make an astonishingly good high fidelity amplifier for use with pick-up. F.M. radio or small P.A. set up. etc The free 40 page manual supplied, details many other applications which this remarkable IC make possible. It is the equivalent of a 22 tran-
sistor circuit contained within a 16 lead DIL package, and the finned heat sink is sufficient for all requirements. The Super IC. 12 is compatibie with Project 60 modules which would be used with the $Z .50$ and $Z .30$ amplifiers. Complete with free manual and printed circuit board.

## SPECIFICATIONS

Output power: 6 watts RMS continuous (12 watts peak). 6-8 . Frequency Response: 5 Hz to $100 \mathrm{KHz} \pm 1 \mathrm{~dB}$. Total Harmonic Oistortion: Less than $1 \%$. (Typical $0.1 \%$ ) at all output powers and frequencies in the audio band (28V). Load Impedance: 3 to 15 ohms. Input Impedance: 250 Kohms nominal. Power Gain: 90 dB (1,000,000,000 times) after feedback. Supply Voltage: 6 to 28 V . Quiescent current: 8 mA at 28 V . Size: $22 \times 45 \times 28 \mathrm{~mm}$ including pins and heat sink..
Manual available separately 150 post free.
With FREE printed circuit board and 40 page manual.
f2.98 Post free

SPECIFICATIONS
Input sensitivities: Radio - up to 3 mV Mag. p.u. 3 mV correct to R.I.A.A. curve +1 dB .20 to 25.000 Hz . Ceramic p.U. - up to 3 mV : Aux -up to 3 mV . Output : 250 mV Signal to noise ratio: better than 70 dB Channel matching: within 1 dB .
Tone controls: TREBLE $\pm 12$ to -12 dB at $10 \mathrm{KHz}:$ BASS +12 to -12 dB at 100 Hz Front panel : brished aluminium with black knobs and controls.
Size : $66 \times 40 \times 207 \mathrm{~mm}$,

## Project 605



The easy way to buy and build

## the world's most advanced high fidelity modules


#### Abstract

Z. 30 \& Z. 50 power amplifiers

The $Z .30$ and $Z .50$ are of advanced design using silicon epitaxial planar transistors to provide unsurpassed standards of performance. Total harmonic distortion is an incredibly low $0.02 \%$ at 15 w ( $8 \Omega$ ) and all lower outputs. Whether you use $Z .30$ or $Z .50$ amplifiers in your Project 60 system will depend on personal preference, but they are the same size and are intended for use principally with other units in the Project 60 range. Therr performance and design are such, however, that $Z .50$ s and $Z .30$ may be used in a far wider range of applications. SPECIFICATIONS ( 2.50 units are interchangeab/e with $Z .30 \mathrm{~s}$ in al/ app/ications). - Power Outputs : Z. 3015 watts R.M.S. into 80 hms using 35 volts : 20 watts R.M.S. into 3 ohms using 30 volts. Z.50 40 watts R.M.S. into 30 hms using 40 volts: 30 watts R.M.S. into 8 ohms using 50 volts.

Frequency response; 30 to $300.000 \mathrm{~Hz} \pm 1 \mathrm{~dB}$. Distortion: $0.02 \%$ into 8 ohms . Signal to noise ratio: better than 70 dB unweighted. Input sensitivity: 250 mV into 100 Kohms (for 15 w into $8 \Omega$ ). For speakers from 3 to is ohms impedance. Size: $14 \times 80 \times 57 \mathrm{~mm}$.


## Project 60 Stereo F.M. Tuner

Buitt and tested. Post free.
The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now. Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other advanced features include varicap diode tuning, printed circuit coils. an 1.C. in the specially designed stero decoder and switchable squelch circuit for silent tuning between stations. In terms of high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically. a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.

(I) $7 \mu \mathrm{~V}$ for lock-in over full deviation. Squelch level: Typically $20 \mu \mathrm{~V}$. Signal to noise ratio: $>65 \mathrm{~dB}$. Audio frequency response: $10 \mathrm{~Hz}-15 \mathrm{KHz}( \pm 1 \mathrm{~dB})$. Total harmonic distortion: $0.15 \%$ for $30 \%$ modulation. Stereo decoder operating level: $2 \mu \mathrm{~V}$. Cross talk: 40 dB . Output voltage: $2 \times 150 \mathrm{mV}$ R.M.S. maximum Operating voltage: $25-30 \mathrm{VDC}$. Indicators: Stereo on; tuning. Size: $93 \times 40 \times 207 \mathrm{~mm}$.


## A.F.U. High \& Low Pass Filter Unit

Built, tested and guaranteed.
$£ 5.98$
For use between Stereo 60 unit and two $Z .30$ s or $Z .50$ s. The unit is very easily mounted and is unique in that the cut-off frequencies are continuously variable. As attenuation in the rejected band is rapid ( 12 dB /octave). there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The A.F.U. is suitable for use with any other amplifier system. There are two filter sections - rumble (high pass) and scratch (low pass). H.F. cut-off (-3dB)


## Power Supply Units

Designed specifically for use with the Project 60 system of your choice. Use PZ. 5 for normal Z.30 assemblies and PZ. 6 or PZ. 8 where a stabilised supply is essential.

PZ. 530 volts unstabilised $\mathbf{£ 4 . 9 8}$

Typical Project 60 applications

| Syster | The Units to use | together with | Units cost |
| :---: | :---: | :---: | :---: |
| Simple battery record player | 2.30 | Crystal P.U.. 12 V battery volume control, etc. | £4.48 |
| Mains powered record player | Z.30, PZ. 5 | Ctystal or ceramic P.U. volume control, etc. | £9.45 |
| 12W. RMS continuous sine wave stereo amp. for average needs | $\begin{aligned} & 2 \times \mathrm{Z} .30 \mathrm{~s} \text {, Stereo } \\ & 60 ; \text { PZ. } 5 \end{aligned}$ | Crystal. ceramic ormag P.U. F.M. Tuner, etc. | £23.90 |
| 25W. RMS continuous sine wave stereo amp. using low efficiency (high performance) speakers | $\begin{aligned} & 2 \times Z .30 \text { s, Stereo } \\ & 60 ; \text { PZ. } 6 \end{aligned}$ | High quality ceramic or magnetıc P.U.. F.M. Tuner. Tape Deck, etc. | £26.90 |
| 80W. (3 ohms) RMS continuous sine wave de luxe stereo amplifier. (60W. RMS into 8 ohms) | $2 \times 2.50$ s, Stereo 60; PZ.8, mains transformer | As above | £34.88 |
| Indoor P.A. | Z.50, PZ.8, mains transformer | Mic., guitar, speakers. etc. controls | £19.43 |
| F.M. Stereo Tuner (£25) \& A.F.U. (£5.98) may be added as required. |  |  |  |

## Guarantee

If, within 3 months of purchasing any product direct from Sinclair Radionice Lid. you are distatisfied with it, your money will be refunded at once. Many Sinclair appointed Stockists aleo offer this same guarentee in co-operation with Sinclair Radionicas Ltd
Each Prolect 60 module is tested before leaving our factory and is guaranted to work perfectly. Should any defect arise in normal use. we will service it at once and without any charge to you, if it is returned within two vears from the date of purchase. Outside this period of guerentee a amall charge (typically f 1.00 ) will be made. No charge is made for postage by surface man. Air Mail is charged at cont.


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Catalogue which contains data sheets for most of the components listed will be sent free on request 10p stamp appreciated.

OPEN ALL DAY SATURDAYS

## RESISTORS

W lokra high stability carbon film-very low noise-capless construction W Mullard CR25 carbon

| Power watts |  |  | Values |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Tolerance | Range | available | 1-99 | $100+$ |
| $\ddagger$ | 5\% | $4.7 n-2 \cdot 2 \mathrm{Mn}$ | E24 | Ip | $0.8 p$ |
| , | 10\% | 3.3Mn-10MR | El2 | 1 p | 0.8 p |
| \% | 2\% | $100-1 \mathrm{Ma}$ | E24 | 3.5p | 3p |
| 1 | 10\% | 10-390 | E12 | $1 p$ | 0.8p |
| $t$ | 5\% | 4.7n-1Mn | E12 | 1 p | 0.8 p |
| 4 | 10\% | $18-100$ | E12 | 6p | $5 \cdot 5 \mathrm{p}$ |

## DEVELOPMENT PACK

0.5 watt $5 \%$ iskra resistors 5 off each value 4.70 to 1 Ma.

FOTENTIOMETERS
Carbon track $5 k \Omega$ to $2 M \Omega, \log$ or linear (log $t W, \operatorname{lin} \ddagger W)$
Carbon track
Single, $12 p$. Dual gang (stereo), 40p. Single D.P. switeh 24p.
SKELETON PRESET POTENTIOMETERS
Linear: $100,250,500 \Omega$ and decades to $5 \mathrm{M} \Omega$. Horizontal or vertical P.C mounting (0.1 matrix),

| TRANSISTORS |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| ACIO7 | 15p | AFI25 | 20p | BDI32 |
| ACI |  |  |  |  |

## BRUSHED ALUMINIUM PANELS

SLIDER POTENTIOMETERS
$86 \mathrm{~mm} \times 9 \mathrm{~mm} \times 16 \mathrm{~mm}$, length of track 59 mm INGLE IOK, 25 K . 100 K log or tin. 40 p
DUAL GANG. IOK $+10 K$ erc. log or tin. 60 p KNOB FOR ABOVE 12p
FRONT PANEL 65p.
with slider pots. Grey or mith slots cut for use plete with fixings for 4 pors

## COMPACT

 CASSETTES—IN BOXC90
C1 20

MULLAND POLYESTER CAPACITORS C296 SERIES
$\begin{array}{lllllll}400 V: & 0.001 \mu F, & 0.0015 \mu F & 0.0022 \mu F, & 0.0033 \mu F, & 0.0047 \mu F, & 21 p,\end{array} 0.0068 \mu \mathrm{~F}, 0.01 \mu \mathrm{~F}$
 $160 \mathrm{~V}: 0.01 \mu \mathrm{~F}, 0.015 \mu \mathrm{~F}, 0.022 \mu \mathrm{~F}, 0.033 \mu \mathrm{~F}, 0.047 \mu \mathrm{~F}, 0.068 \mu \mathrm{~F}, 3 \mathrm{p} .0 .1 \mu \mathrm{~F} 3 \frac{1}{2} \mathrm{p} .0 .15 \mu \mathrm{~F}$ 4 $\frac{1}{2} \mathrm{p}$ $0.22 \mu \mathrm{~F}, 5 \mathrm{p} .0 .33 \mu \mathrm{~F}, 6 \mathrm{p}$. $0.47 \mu \mathrm{~F}, 7 \mathrm{p}$. $0.68 \mu \mathrm{~F}$, $11 \mathrm{p} .1 .0 \mu \mathrm{~F}$, 13 p
MULLARD POLYESTER CAPACITORS C2B0 SERIES
$250 V$ P. $31 \mathrm{p} .0 .1 \mu \mathrm{~F}, 4 \mathrm{p}, 0.15 \mu \mathrm{~F}, 0.22 \mu \mathrm{~F}, 5 \mathrm{p}, 0.33 \mu \mathrm{~F}, 61 \mathrm{p} .0 .47 \mu \mathrm{~F}, 81 \mathrm{p}, 0.68 \mu \mathrm{~F}, 11 \mathrm{p} .1 .0 \mu \mathrm{~F}, 13 \mathrm{p}$ $1 \cdot 5 \mu \mathrm{~F}, 20 \mathrm{p} .2 \cdot 2 \mu \mathrm{~F}, 24 \mathrm{p}$.
MYLAR FILM CAPACHTORS IOOV CERAMIC DISC CAPACITORS $0.001 \mu \mathrm{~F}, \quad 0.002 \mu \mathrm{~F}, \quad 0.005 \mu \mathrm{~F}, \quad 0 \mathrm{O} 0 \mathrm{I} \mu \mathrm{F}, \quad 0.02 \mu \mathrm{~F}$ 24p. $0.04 \mu \mathrm{~F}, 0.05 \mu \mathrm{~F}, 0.068 \mu \mathrm{~F}, 0.1 \mu \mathrm{~F}, 31 \mathrm{p}$.

## ELECTROLYTIC CAPACITORS-MULLARD OI5/6/7 RANGE REPLACES C426

 C457RANGES( $\mu$ F/v) $1.0 / 63,15 / 63,22 / 63,3 \cdot 3 / 63,47 / 63,6 \cdot 8 / 40,10 / 25,10 / 63,15 / 16,15 / 40,15 / 63$ $100 / 10$ / $00 / 25$, $100 / 40$ 150/6.3 $150 / 16$, $150 / 25$ 220/4 $220 / 10,220 / 16 \quad 330 / 4,330 / 10$ $100 / 10,100 / 25,100 / 40,150 / 6 \cdot 3,150 / 16,130 / 25,220 / 4,220 / 10,220 / 16,330 / 4,330 / 10$
$470 / 6 \cdot 3,5 p$ each. $68 / 63,150 / 40,220 / 25,330 / 16,470 / 10,680 / 6 \cdot 3,1.000 / 4,9 p .100 / 63$ $150 / 63,220 / 40,470 / 25,680 / 16,1.000 / 10,1,500 / 6 \cdot 3,12 \mathrm{p} .220 / 63,470 / 40,680 / 25$ $1.000 / 16,1,500 / 10,2,200 / 6 \cdot 3,15 ; \quad 330 / 63,680 / 40,1,000 / 25,1,500 / 16,2,200 / 10$ 3,300/6.3. 4,700/4, 18p.

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2 pin, 3 pin, 5 pin $180^{\circ}, 5$ pin $240^{\circ}, 6$ pin

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Plug 12 p . Socket 8 p.
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$\begin{array}{lllll}0.022 \mu \mathrm{~F} & 12 \mathrm{p} & 0.047 \mu \mathrm{~F} & 13 \mathrm{p} & 0.22 \mu \mathrm{~F}\end{array} \quad$ 20p
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Wormal Price $[39.60$ luncis (see Olson AA-310) and tape decks. And it's pot out. Iuners (set Olion RA- 310 ) and tape decks. And it's got out.
Dutif for taping and for heodphonas. Thera ate separate bass and treble comtrols, sepssate left and Right chanmal volume trable notes when listaning at low output lavals. Fresuancy inta anme ingun $\mathrm{Hz} \pm 38 \mathrm{~A}$. Dutput: 20 watts r.m.s. por channs
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## P

8-1AACK STEREO CARTRIDG TAPE DECKGMOOEL AP-1000S The popula ear-Jet type record ing unit is the Heart of the fantastic and playbat facilities. Automatic rack change hh manual over-ride Aight mic sackets on front panel. Dual recording level meter, and Leit and Right volume comioks. Built-in pre-amp. Tape speed 3 a ips ( 9.5 cms ). Frequency response, playback $30-10.000 \mathrm{~Hz}$; recording/playback $30-8.000 \mathrm{~Hz}$. Line output: fully variable 0.500 mV . The Size $16^{\prime \prime}$ wide. $4^{\prime \prime}$ high, $9^{\prime \prime}$ deep. Cabinet in walnut. Including connecting leads.

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and tuning controis with
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STEREOAMPLIFIEA Here's fabulous. exciting value in miniature! This high quality stereo amplifier measures only 9 " wide $\times 3$
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