PRACTICAL

FRONI

JANUARY 1975

25p

FOR SCINTILLATING DECORATION

SCINTILLATING SOUND



HI-FI STEREO AMPLIFIER



TOTAL BUILDING COSTS

£7-23 P.P. & INS. 44p (Overseas P.P.

NEW EDU-KIT MAJOR

COMPLETELY SOLDERLESS ELECTRONIC CONSTRUCTION KIT BUILD THESE PROIECTS WITHOUT SOLDERING IRON OR SOLDER

- 4 Transistor Earpiece Radio
- Signal Tracer

Amplifier

Components include:

- Signal Injector
- Transistor Tester NPN
- -PNP
 4 Transistor Push Pull
 Amplifier
 5 Transistor Push Pull
- 7 Transistor Loud-speaker Radio MW/LW.
- Transistor Short Wave Radio
- Electronic Metronome
 Electronic Noise Genera-
- tor Batteryless Crystal Radio One Transistor Radio

24 Resistors ● 21 Capacitors ● 10 Transistors ● 3½ Loudspeaker ● Earpiece ● Mica Basehoard
 312-way Connectors ● 2 Volume Controls ● 2 Slider Switches ● 1 Tuning Condenser ● 3 Knobs
 Ready Wound MW/LW/SW Coils ● Ferrite Rod ● 6½ yards of wire ● 1 yard of sleeving, etc.
 Parts price list and plans 50p (free with parts)

- Transistor Regenera-
- tive Radio
 3 Transistor Regenerative Radio
 Audible Continuity
 Tester
- · Sensitive Pre-Amplifier

NEW ROAMER 🗥 NINE

WITH V.H.F. INCLUDING **AIRCRAFT**

Nine Tran-sistors, 9 Tunable wavebande

Roamer Ten. Built in ferrite rod aerial for MW/LW in fertite rod aerial for MW/LW. Retractable chrome plated telescopic aerial for VHF and SW. Push Pull output using 600 mW transistors. 9 Transistors and 3 diodes, tuning condenser with VHF section, separate coil for aircraft, moving coil loudspeaker, volume ON/OFF and wavechangecontrols. Attractive all white case with red grille and carrying strap. Size 9/in × 7in × 2½m approx. Parts price list and plans 30p (FREE with parts).

TOTAL BUILDING £6.95 (+8% VAT 55p)

(OVERSEAS P. & P. £1-85)

POCKET FIVE

NOW WITH 3" LOUDSPEAKER 3 Tunable wavebands MW/LW and Trawler Band, 7 stages. Trawlet

rand. 7 stages, transistoria and diodes, supersensitive ferrite rod aerial, attractive Black and Gold Case. Size 5\(\frac{1}{2}\) in X 1\(\frac{1}{2}\) in 2 approx. Plans and parts price 1st 15\(\frac{1}{2}\) (+8\(\frac{1}{2}\) VAT 20\(\frac{1}{2}\)) (FREE with parts)



Total Building Costs £2.75 P.P. & Ins.

(+8% VAT 21p) (Overseas P. & P. £1-25) (parts).

(+8% VAT 20p) P.P. & Ins. 26p (Overseas P. P. £1.25)

TRANSONA FIVE NOW WITH 3" LOUDSPEAKER

Wavebands, transistors and speaker as Pocket Five. Larger Case with Red Speaker Grille and Tuning Dial. Plans and parts price list 15p (FREE with parts)

NEW EVERYDAY SERIES

Build this exciting New series of designs

EV5 5 Transistors and 2 diodes. MW/LW. Powered by 4½ volt Battery. Ferrite rod aerial, tuning condenser, volume control, and now with 3 londspeaker. Attractive case with red speaker grille. Size 9in × 54in × 25in approx. Parts price list and plans 15p (FREE with parts).

TOTAL
BUILDING
\$\frac{\text{4.9°}}{\text{4.9°}}\text{VAT 23p}\$

EV6 Case and looks as above
6 Transistors and 3 diodes. Powered by 9 volt Battery, Perrite roll agrid 3 did 1 and standards at 1 and 1 ovil Battery. Ferrite rod aerial, 3" loudes. Powere to ye wort Battery. Ferrite rod aerial, 3" loudspeaker, etc., MW/LW coverage. Push Pull Output. Parts price list and plans 15p (FREE with parts.)

TOTAL

BUILDING
COSTS

OF LATRIDAT

P. & P. £1:25)

(+8% VAT 29p)

P.P. & INS. 30p (OVERSEAS P. & P. £1:25)

ROAMER EIGHT Mk. I

NOW WITH VARIABLE TONE CONTROL

7 TUNABLE WAVEBANDS:

7 TURABLE WAYEBANDS:
MW1, MW2, LW, SW1,
SW2, SW3 AND TRAWLER BAND, Built-in ferrite rod
aerial for MW and LW. Chrome plated telescopic
aerial can be angled and rotated for peak shortwave listening. Push-pull output using 600mW
transistors. Car aerial and tape record sockets.
Selectivity switch. Stransistors plus 3 diodes. Latest 47
wat L'Errite Manuel Justenecker. Air senger designed. Selectivity switch. S transistors plus 3 (todes. Latest 4' 2 watt Ferrite Magnet loudspeaker. Air spaced ganged tuning condenser. Volume/on/off, tuning, wave change and tone controls. Attractive case in rich chestnut shade with gold blocking. Size 9 in x 7 in x 4 in approx. Easy to follow instructions and diagrams. Parts price list and plans 25p (FREE with parts).

BUILDING

£6.98

(OVERSEAS P. & P. £1-85)

ROAMER WITH VHF

INCLUDING AIRCRAFT

AIRCRAFT

10 TRANSISTORS.
9 TUNABLE
WAVE BANDS.
MWI, MW2, LW.
SWI, SW2, SW3,
TRAWLER BAND,
VHF AND LOCAL
STATIONS. ALSO AIRCRAFT BAND
Latest 4" 2 watt Perrite Magnet Loudspeaker.
Built-in ferrite rod aerial for MW/LW.
Chroine plated 7 section telescopic aerial, can be angled and rotated for peak short wave and VIIF listening. Push-pull output using 600mW transistors pius 3 diodes. Ganged tuning condenser with VIHF section. Section. Separate coil for Aircraft Band. Volume/on/off, wave change and tone controls. Attractive case in black with silver blocking. Size 9m x 7in x 4in. Easy to follow instructions and diagrams. Parts price list and plans 30p (PREE with parts).

TOTAL

P.P. & INS. 52p

TOTAL BUILDING

£8.50

(OVERSEAS P. & P. £1-85)



Components include: Tuning Condenser: 2 Volume Controls: 2 Slider Switches: Fine tone 3" moving coil Speaker: Terminal Strip: Ferrite Rod Aerial: Battery Clips: 4 Tag Boards: 10 Transistors: 4 Diodes: Resistors: Capaciters: Three jin Knobs. Units once constructed are detachable from Master Unit, enabling thom, to be accept for thurne use. [Jeal for Schools them to be stored for future use. Ideal for Schools. Educational Authorities and all those interested in radio construction. Parts price list and plans 25p (FREE with parts).

TOTAL BUILDING COSTS

£5.50 (+8% VAT 44p)

P.P. & INS. 33p (OVERSEAS P. & P. £1-85)

PF1

TRANS EIGHT 8 TRANSISTORS AND 3 DIODES

6 TUNABLE WAVEBANDS, MW, LW, SWI, SW2, SW3 AND TRAWLER BAND. Sensitive ferrite rod acrial for MW, and LW. Telescopic aerial for short waves. 3in speaker. 8 improved type transistors plus 2 diodes. Attractive case in black with red grille, dial and black knobs with polished metal inserts. Size 9in × 51in × 21in approx. Pushpull output. Battery economiser switch for extended battery life. Ample power to drive a larger speaker Parts. price list and plans 25p (FREE with parts). TOTAL BUILDING COSTS

£4*48 (OVERSEAS P. & P. £1·25)

ROAMER SIX CASE AND LOOKS AS TRANS EIGHT 6 TUNABLE WAVEBANDS: MW, LW, SW1, SW2, TRAWLER BAND PLUS AN EXTRA MW BAND FOR EASIER TUNING OF LUXEMBOURG, ETC. Sensitive ferrite rod aerial and telescopic aerial for short waves. 3 in speaker. 8 stages—6 transistors and 2 diodes, etc. Attractive black case with red grille, dial and black knobs with polished metal inserts. Size 9 in x 5 in x 2 fin a y 2 in x 2 fin x 5 in x 5

£3.98	P.P. & IN. 31p (OVERSEAS P. & P. £1:85)
(+8% VAT 32p)	,

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ELECTRONICS

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Our February issue will be published on Friday, January 10, 1975

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PLINTH AND Garrard SP25 Mk IV deck Goldring G800 Cartridge Teak fin-shed Plinth Cover hinged).

GLOBAL'S PRICE £19.95

Carr & Ins £1 93

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Garrard SP25 Mk IV Chassis Garrard 86SB P C Carl (Mod) Garrard 86SB C Chassis Garrard 401 Chassis Goldring 101 Mk II P C 800 Goldring GL75 P C 6800 Goldring GL75 P C 6800 Goldring GL75 P C 700 Coldring GL75 P C 700 C 7 £12-95 £47-50 £22-40 £33-25 £23-70 £39-95 £44-40 £66-95 P.O.A. Sansui SR 212 Thorans TD125 Mk II £73 - 75

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Please add £1-05 for P. & P. & Ins.	
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Goodmans Meto 3SL
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Amstrad IC2000 Mk. II with increased power 25 ± 25W amplifier. Complete with a pair of Amstrad Acoustra 2500 speakers. Garrard SP25 Mk. IV deck. 6800 Cart. Plinth Cover (non hinged). All leads.

GLOBAL'S PRICE £80.75 Carr. & Ins £3-30

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SIEREO READPRONES	
Please add 42p for P. & P. & Ins.	
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Koss ESP 9	£65.75
Koss K7/11 Red Devil	£10 - 99
Koss 6	£11-70
Koss K6/LC	£13 - 25
Koss KD 727E	£14-90
Koss 747	£19 - 75
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Kosa HV1/LC/A N/P	£24-80
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Bargain pack PEP2A, £4.90.

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TMK 200 MULTIMETER KIT

Build yourself a quality 20000 opv. quality 20000 opv.
multimeter and
seve money.
Complete kit with
meter scale, meter scale, movement and rotary range selector ready mounted in cabinet. All parts,

cabinet. All parts, batteries, test prode and batteries, test prode and batteries, test prode and productions. Bangaes: 0/0.6/5/30/120 120/600/1200 V D.C. 0/6/30/120 500/1200 V A.C. Current: 0/0.8/6 0/600m. Resistance: 0/10/100K/1/10 Meg ohms. Decibels: —20 to +63db. Size: 90 x 150 x 35mm.

OUR PRICE £7,95 P&P 30p.

AUDIOTRONIC Model ATM1

AUDIOTRONIC Mo
Top value 1,000
opv pocket multimeter, Ranges: —
0/10/60/250/1,000
volt AC and DC.
DC current 0-1mA/
100mA, Resistences
0/150k ohms.
Decibets: —10 to +22dB, Size 90 ×
60 × 28mm.
Complete with test leads.

OUR PRICE £3.25

P&P 15p

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AUDIOTRONIC Model ATM5

AUDIOTRONIC Mode
Jawel movement,
attractively moulded
case with adaysise
ohms adjustment.
Ranges: 0.31/6/150/
300/1200V AC,
12500 opp), 0-300
uA/0-300mA DC,
Ranges of the second of the

OUR PRICE £3.95 P & P 20p

MODEL TH12

MUDEL ITII2
20,000 opv. Overload
protection. Slide switch
selector. 0/0.25/2.5/10/
50/150/1000V DC. 0/10/
50/250/1000V AC. 0/
50uA/25/250mA DC,
0/3k/30k/300k/3 Megohr
+50dB. hms -20 to

OUR PRICE £5.95 P&P 30p

HIOKI 720X VOM A versatile, securate measuring instrument. 20,000 opv. 0/ 5/25/100/500/ 1000V DC. 0/10,50/250/1000V AC. 0–50uA/ 250mA. 0–20k/ 2 Meaohuse.

2 Megohms. OUR PRICE P&P 30p

MODEL PL436 20,000 opv DC. 8000 opv AC. Mirror scale -6/3/12/30/120/ 600V DC. 3/30/ 120/600V DC. 50/600µA/60/

10/100K/1 Meg/10 Meg Dhm. **OUR PRICE £6.97** P&P30p. OUR PRICE £13.50

U4323 MULTIMETER

U4323 MULTIMETER
20,000,00v. Simple
unit with audio/IF
cosilitator. Suitable
for general receiver
runing. Ranges:
0.5/2.5/10/60/250/
500/1000V DC.
2.5/10/15/250/500/A DC. Resistance:
x10, x 100, x 1,000, x 10,000 (500.
5000, 5k0, 5k0 centre scale) 500Ω, 5kΩ, 50kΩ centre scale)
Battery operated. Size: 160 x 97 x
40mm, Supplied in carrying case complete with test leads.

OUR PRICE £7.70 P&P 30p

HIOKI 730X

30,000 opv. Over-load protection. 8/30/60/300/600/ 1200V DC. 12/60/ 120/600/1200V AC. 60/μA/ 30mA/300mA. 2K/200K Meg Dhm.

OUR PRICE £7.50 P&P 30p.

U4324 MULTIMETER

U4324 MULTIMETER
High sensitivity, overload protectived-oper10.6/1-2/3/12/30
60/120/600/1200
DC. 3/6/15/60/150
00/000/600/000
DC. 3/6/15/60/150
00/600/6000/AC
0.3/3/30/300mA/
3A AC. Resistence:
25/500 ohns/0.5/550/500k ohns/5
Mohms. Decibels: -10 to +12.dis. Size
Homs. Decibels: -10 to +12.dis. Size
test with sense, spare diode and
instructions.

OUR PRICE £9.25

U435 MULTIMETER

20,000opv. Ranges:

20,000ov. Ranges: 75mV/2.5/10/25/ 100/250/500/1000v DC C. 25/10/25/100/ AC C5/100n AC S6/100n AC S6 **OUR PRICE F8.75** P&P 30p

U4312 MULTIMETER

U4312 MULTIMETER
extremely sturdy
instrument local
organization
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OUR PRICE £10.25 P&P 50p

U91 Clamp VOLT

AMMETER remete LEN For measuring AC voltage and current without breaking circuit, Ranges: 300/600V AC. Current: 10/25/100/250/500A. Accuracy 4%. Size 283 x 94 x 36mm. Complete with carrying case, leads and fuses.

P&P 30p

MODEL 500

30,000 opv with overload protect tion. Mirror scale Overnose protect tion. Mirror scale, 0/0.5/2.5/10/25/ 100/250/500/ 1000V DC. 0/2.5/10/25/100/ 250/500/1000V AC, 0/50u.A/5/50/ 500mA, 12A DC. 0/60k/6 meg/60 megohms.

DUR PRICE £13.95 Carr, paid

HIOKI 750X VOLT-OHM-MILLIAMETER

OUR PRICE £11.95

TMK MODEL TW50K

TMK MUULL I WDUN
46 ranges, mirror
scale, 50k/V DC
50k/V AC,
DC Volts: 0.125/
00/150/, AC Volts
1,5/3/5/10/25/50/
156/250/50/00mA/5/
10A, Resistence:
10k/100k/1 Meg/
10 Meg ohms. -20 to +81.5d8,

OUR PRICE £12.50 P&P 20n

HIOKI MODEL 700X

HIDKI MODEL 70D)
100,000opv, Overload
protection, Mirror scale,
0.3/0.6/1.2/1.5/3/6/
12/30/60/120J20/
600/1200V DC,
1.5/3/6/12/30/60/150/
300/800/1200V AC,
15/3/60/12/30/60/150/
150/500mA/6/12A DC,
24/2006/24W/20MOhms,
-20 to +63dB,

OUR PRICE £14.95

Model HT100B4 MULTIMETER

MODEL C7202EN

20,000 o.p.v. DC. 10,000 o.p.v. AC Mirror Scale. 5/25/50/250/500/ 1000/2500 V. DC. 10/50/100/500/1000 V. AC. DC Resistance V. AC. DC Hesistand
×10, ×1000 (30Ω
centre scale) DC
Current 50uA/
2·5mA/250mA. —20
to +68 dB,



OUR PRICE £6.50 P&P 30g

KAMODEN HM720B FET VDM AMRUUEN HM/2 Input impedence 10 Megohms. Ranges:— 0/25/1/2.5/10/50/ 1000V DC. 0/2.5/10 50/250/1000V AC. 0/25uA/2.5/25/250 mA DC. 0/5k/50k/500k/5 M 500 Megohms **OUR PRICE**

£21.00 P&P40p KAMODEN 360 MULTIMETER

RAMODEN 360 MULTIMETER
High sensitives
High sensitives
CA 104chm/V
5" mirror scale,
overload protected, Ranges: 0.5/
25/10/50/250/
109/250/000/
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OUR PRICE £17.50 P&P40p

TMK MOOEL 117 FET

ELECTRONIC VOLTMETER

ELECTRONIC VOLIMETER
Battery operated.
11 Meg input, 26
ranges, Large 43"
mitror scale, Size:
149x 117x 60mm,
0.3–12000V DC,
3–300V RWS AC,
8–800V P.P.
DC current 0.12–
12mA, Resistence
up to 2000MOhms, Decibels: –20 to
451dB, Supplied complete with leads
and instructions.

OUR PRICE £18.50 P&P 20p TMK 100K LAB TESTER

TMK 100K LAB TESTER
100,0000pv, 65"
seale. Buzzer
short circuit check,
Sensitivity, 100,000
pv DC, Sk/V AS,
10/50/250/1000V
Cc. 34/10/50/250/
10/50/250/1000V
DC,
current 10/100_ARsintance:
10/100/2.5/10A. Resintance:
10/100/2.

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370WTR MULTIMETER

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High sensitivity tester. 200,000 opv Overload protected. Mirror scale. Overload protected, Mirror scale, Ranges: -0/.06/.3 3/30/120/800/ 1200V DC, 0/3 12/60/300/11200 V AC, 0/6uA/ 1.2mA/120mA/ 600mA/12A DC 0/12A AC, -20 to +63dB, 0/2k/200k/ 2 Meg/200 Megohm

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MODEL AS:100D VOM
100.000 opv.
Mirror scale.
Built-in meter protection. 0/3/
12/60/120/300/
600/1200V DC.
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600/1200V DC.
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MODEL C7208FM

30,000 opv DC. 15,000 opv AC. 6/3/15/60/300/600/ 1200 V. DC. 6/30/ 120/600/1200 V. AC. DC Resistance x1, x10, x100, x1000 (50Ω centre scale) DC Current 30uA/ 3/30/600mA.—20 to +63d8.



OUR PRICE £8.95 P&P30p

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MODEL U4311 Sub-standard
Multi-range Volt-Ammeter
Sensitivity 330
Ohms/Volt AC
and DC.
Accuracy 0.5%
ACC. 1% AC.
Scale length:
165mm.
0/300/750.04
1030/75150300/750mV/1.5/3/7.5/15/
30/75/150/300/750W DC. 0/75mmV/
1.5/3/7.5/15/300/750W DC. 0/75mmV/
1.5/3/7.5/15/300/750W DC. 0/75mmV/
AC. Automatic cut out device. Supplied complete with vest least, manual AC. Automatic cut out device. Supplied complete with test leads, manua and test certificates.

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50,000 opv. Mirror scale. Meter protection. 0/-3/3/12/60/120/ 300/600/1200V DC 300/800/1200V DC. 0/8/30/120/, 300/800/1200V DC.

300/800/1200V DC. 0/30µA/8/ 80/300 mA/ 12 Amp. 0/10K/ 1m/10m/100 Meg Ohms. – 20 to +17 dB. OUR PRICE £12.50 P&P30p.

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TRANSITOR
High quality
instrument to
test reverse leak
current and DC
current. Amplification factor of
NPN, PNP, diodes,
transistors, SCR's
clear scale meter.
Operates from
internal batteries.
Complete with
instructions, leads
carrying handle.
OHR PRICE £1.

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U4341 Multimeter & Transistor Tester
Zyranges. 16,700opv.
Overload protected,
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OUR PRICE £10.50 P&P 30p

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TRANSISTOR TESTER
100,000-pv. Mirror
scale. Overoed
rortection. 0/0 12/
0.673/12/30/120/
0.600 D.C. 0/6/30/
120/600V AC. 0/12/6000 A/12/
300mA/6/12A DC
0/10/6/1 Meg/
100 Meg.
-20 to +5od8.
0.01-0.2 MFD
and ICO. Complete with instructions, batteries and leads.

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Handy SWR meter for transmitter antenna align-ment, with bullt-in field strength meter. Accuracy 5%, Impedence 52' Indic-ator 100uA DC. Full scale 5 section collapsible section collapsible na. Size 145 x 50 x



OUR PRICE £4.25

P&P 30n

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For display of pulsed
and periodic waveforms in electronic
circuits. VERT. AMP.
Bandwidth. 10MHz.
Sensitivity at 100kHz
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Presst triggered sweep
1—3000usec. Free running 20—200
kHz in nine ranges. Calibrator pigs.
220 x 360 x 430mm. 115—230V AC.
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RUSSIAN CI16 Double Beam OSCILLOSCOPE

Separate Y1 and Y2 amplifiers. Rectangular 5" x 4" CRT. Calibrated triggered sweep from 0.2usec, to 100 milli-sec/cm. Free running time base, 50Hz-1MHz. Built-in time base Calibrator and amp Supplied complete and incomp plitude Calibrator

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MODEL TE15 GRID DIP METER Transistorised. Oper ates as Grid Dip, Oscillator, Absorb-tion Wave Meter and Oscillating Detector. Frequency range 440kHz – 280MHz in six coils. 500u a meter. 9V battery operation. Size: 180 x 80 x 40mm

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A new portable bridge offsring accuracy at low cost. Resistance: 6 ranges: 0.1 ohm-11.1 megahm ± 1% Inductance: 6 ranges: 10 foranges: 10 foranges: 10 foranges: 10 foranges: 10 foranges: 11/1000-111100 ± 1% Bridge Voltage at 1,000cps. Operated from 9-voit battery. 100 microamp meter Indication. Size 7% x 5" x 2" m. Dager access the size of amp meter Indication.
5" x 2" OUR PRICE £25.00 P&P 30P

TE-20D RF SIGNAL GENERATOR

. Accurate wide range covering 120 kHz-500 MHz on 6 bands.
Directly calibrated.
Variable R.F. signal generato

Variable H.F. audio output, Xtal soc attenuator audio output, Xtal soc for calibration. 220/240V a.c. Brand new with instructions.
Size 140mm x 215mm x 170mm OUR PRICE £17.50 P& P 50p

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Sine 20cps to 200kHz on 4 bands. Square 20 cps to 30 kHz. Output impedence 5000 Ohms. 200/250V AC operatio

AC operation. Supplied brand guaranteed, with instruction me

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ARF 300 AF/RF SIGNAL

GENERATOR GENERATOR
All transistorised
compact fully
portable. AF sinewave 18Hz to 220
kHz. AF square
wave 18Hz to 100k
Hz. Output Square
Sine wave 10V.
P.P. RF 100kHz to
200MHz. Output
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1V maximum. 220/240V AC operation. Complete with instructions and leads. OUR PRICE £37.50 P&P 50p WALKIE TALKIES

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SKYFON NV7 Super low cost transmitter/ receivers, 100MW with call buzzer and on/off volume control. 7 transistors. Telescopic rod

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MODEL MG 100 SINE SQUARE **WAVE AUDIO** GENERATOR Range 19-220,000Hz Sine

Wave 19—100,000 Hz Square Wave.
Output Sine or Square wave 10v. P. to F
Size 180 x 90 x 90mm.Operation 220/240v. A.C. DUR PRICE £19.95

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High quality ceramic construction. Wind-ings embedded in vitreous enamel. Heavy duty brush wiper. Continuous rating.

Single hole fixing. %" diameter shafts. Bulk quantities available. 25 WATT 10/25/50/100/500/1000/

£1.15 P&P 10p 2500 ohms. 50 WATT 10/50/100/250/500/ 1500/5000 ohms. £1.62 P&P 10p

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£2.34 P&P 15p EMI LOUDSPEAKERS Model 350 13 x 8" with single tweeter/crossover. 20–20,000Hz. 15 watts RMS. Available 8 o

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Heavy duty, ideal for Hi-Fi P.A. Group. OUR PRICE £12.50. P&P 50p.

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Solid state, Veriable
output 5—20V DC
up to 2 Amp. Indepandent meters to
monitor voltage and
current, Output
220/240V AC.
Size: 190 x 136 x

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Beautifully made and finished in two tone ivory/buff, the LE-102A is useful in the home, office or shop and is suitable for use as baby alarm. Wall or desk mounting 57mm speaker/mic gives clear 2-way communication with on/off and volume control on master. and volume control on master unit. Operates on 9V batt. Approx. 60ft lead.

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Will play 8 track stereo CHE cartridge monaurally Channel selector switch. Covers

nedium and long wave bands. Volume and too Earphone socket, Battery/Mains

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EA41 REVERBERATION

AMPLIFIER Self contained

transistorised, battery operated, Simply plug in microphone, guitar atc. and output to your amplifier. Volume control and depth of reverberation control. Beauwalnut cabinet. 184 x 77 x 108mm.

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DHO2S STERED HEADPHONES Wonderful value and excellent performance combined. Adjust able head band. able head band. Impedence 8 oh 20-12,000Hz. OUR PRICE £2.25

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Low cost with excellent response. Foam
rubber earcups. Adjustable headband. 8 ohms
impedence. Frequency
response 25Hz-18kHz.
Complete with cable
and sterao Jack plug.

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P&P 300 **BH001 HEADSET and Boom Vicrophone** Moving coil, ideal for language teaching,

cations etc impedence 16 ohms. Mic-pedence 200 ohms. OUR PRICE £5.95 P&P 30p

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Covers Medium and FM wave bands Slider

bands. Silder volume and tone controls. Battery/Maina operation. Will record direct from radio or through built in condenser micropho plete with batteries, earni and cassette

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OUR PRICE £24-30 P & P 50n

SPECIAL BARGAIN !! STEREOSOUND SPEAKERS

Matched pair of stereo bookshelf speakers. Deluxe teak veneered finish. Size: 368 x 229 x 190mm. 8 ohms. 8 watts Pak. Complete with Din lead.



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FM TUNER CHASSIS

high quality tuner, Size only 153 x 101 x 63mm 3 IF stages. Double tuned

Ample output to feed most amplifiers.
Operates on 9V battery. Covers 88—
108MHz. Ready built, ready for use.

OUR PRICE £8.95

SPECIAL OFFER! SAVE OVER 50%



AMSTRAD 8000/2 Stereo amplifier AMSTHAD 8000/2 Stereo amplifier 7 watts per channel rms. Inputs for tuner tape, phono, Headphone socket. List price £29.95. OUR PRICE £12.95 P&P 60p

SPECIAL OFFER! CONVERT YOUR STEREO SYSTEM



Exclusive offer of GOODWIN 4-CHANNEL CONVERTER and a pair of AD15 10 watt 8 ohm bookshelf speakers enables you to add 4D sound to your existing system. Complete with slimple connection details. Normal retail value £25,50. OUR PRICE £15.80 P& P £1.

GOODWIN CONVERTER available separately £3.95 P & P 50p.

FM TUNER 6 transistor high quality unit— 3 IF stages and double tuned

double tuned discriminator. For use with most amplifiers. Covers 88—108MHz, Powered by 9V battery. OUR PRICE £13.50

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MINIATURE ORGAN MUSIC MASTER AM100



This instrument will give hours of enjoyment will give hours of enjoyment to all the family. Beautifully finished. The Keyboard range can be adjusted to be in tune with any instrument. Operates from internal 9V battery. Fitted with on/off switch, vibrato switch, earphone socket and external 8V D.C. socket.

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numbers for hours minutes A.C. 240V operation. Size approx. 61 x 3 x 31 inches. OUR PRICE £4.50 P&P 50p.

SINCLAIR ICI2 INTEGRATED CIRCUIT AMPLIFIER 31111 complete with

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SINCLAIR Project 80 Modules 240 Power Amp.
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For balancing and gain selection of loudspeakers with additional facility for stereo headphone 0000 switching. Two gain controls, speakers on-off slide

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P&P 3p each. 10 and over Post Free MP7 MIXER-PREAMPLIFIER

MP7 MIXEM-PREAMFLIFTEN
5 Microphone
inputs each with
individual gain
controls enabling
facilities. Battery operated. Size: 235
x127 x 76mm. Inputs: Miss. 3 x 3mV
50k; 2x 3mV 60k obms. Phono. Mag.
4mV 50k; Phono Ceramic 100mV 1
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Stereo III.
All silicon, transistor cransistor amplifier operates from magnetic, ceramic or tuner

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OUR PRICE £8.50 -Ww-

JOSTY KILL



WK ARE APPOINTED STOCKISTS AT Oxford Street, 42 & 34 Lisle Street, 152. Fleet Street, 311 Edgware Road, CROYDON BIRMINGHAM KINGSTON LEICESTER NORTHAMPTON SOUTHEND TUNBRIDGE WELLS WOLVERHAMPTON branches, or by Mail Order.

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USED EXTENSIVELY BY INDUSTRY, GOVERNMENT DEPARTMENTS, EDUCATIONAL AUTHORITIES ETC. Over 200 ranges in stock—other ranges to order. Quantity discounts available. Send for fully illustrated brochure.

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1mA	£3.65		5500
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50mA	£3.65	10V DC	£3.65
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1A DC	£3.65	300V DC	£3.65
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EDGWISE MODEL PE70 Size: 90 x 34mm

50uA	**		£4.15
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A range of high quality moving coil instruments ideal for school experi-ments and other bench applications. 3" mirror scale. The meter move-ment is easily accessible to demonstrate internal working.



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Size: 120	x 1	10mr	n
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100u A			£5.40
200u A			€5.35
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50-0-50u A			£5.40
100-0-100			15.35
500-0-500		**	£5.20
1mA			
		**	£6.20
1-0-1mA	**		£5.20
5mA	**		£5.20
10mA			£5.20
50m A			£6.20
100mA			£5.20
500mA			₹5.20
1A DC			€5.20
5A DC			₹5.20
15A DC			£5.20
	••		
30A DC	••		£5.40
10V DC			£5.20
SOA DC			£5.20
50V DC			£5.20
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Size: 110	x E	33m		
50u A			£4.3	
100uA			£4.2	
200u A			£4.2	
500u A			£4.1	
50-0-50u			£4.2	
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1mA			£4.1	
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10mA			£4.1	
50mA			£4.1	
100mA		**	£4.1	

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10V DC			£4.10
20V DC 50V DC			£4.10 £4.10
300V DC			£4.10

£4.10 £4.10 £4.10 £4.10 1A DC 5A DC 10A DC 5V DC 15V AC 300V AC VU Meter CLEAR PLASTIC MODEL MR 45P

Size: 50 >	50	mn	n
50uA			£3.20
100u A		**	£3.15
200u A			£3.10
			£3.00
50-0-50u			£3.15
100-0-10	0u A	١	£3.10
500-0-50	Ou A	١	£2.95
1mA			€2.95
5m A			£2.95
10m A			£2.95
50mA			€2.95
100mA			£2.95
500mA			£2.95
1A DC			€2.95
5A DC			£2.95
10V DC			£2.95
20V DC			£2.95
50V DC			£2.95
300V DC			€2.95
15V AC	,.		£3.05
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5	300V AC				£3.05
5	S Meter 1	m/	۱		£2.95
5	VU Meter	۲.,			£3.40
5	1A AC 5A AC	**			£2.95
5	10A AC			٠	€2.95
5	20A AC	,.		*	€2.95
5	30A AC	**		*	£2.95

CLEAR PLASTIC MODEL MR 38P

3146. 44 V	72		
50uA			£3.10
100u A	**		£3.05
200u A			£3.00
			£2.88
50-0-50u	Α.		€3.08
100-0-100	λuΑ	٠	£3.00
500-0-500	Du A	١	£2.80
1mA			£2.80
1-0-1mA			£2.80
2mA			£2.80
5mA			£2.80
10mA			€2.80
20mA			£2.80
50mA			£2.80
100mA			£2.80
150mA			£2.80
200mA			£2.80
300mA			£2 80
500mA			£2.80
750mA			€.28
1A DC			€2.8
2A DC			₹2.8
5A DC	**	••	£2.8
10A DC	**		€2.8
3V DC	••	**	£2.8
	••		
10V DC	••	••	£2.8
15V DC			€28



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.80	20V DC	 €2.80
.80	50V DC	 £2.80
.80	100V DC	 £2.80
.80	150V DC	 £2.80
80	300V DC	 £2.85
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.80	50V AC	 £2.90
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RO	S Meter 1mA	£2.80
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80	VU Meter	 £3.20

CLEAR PLASTIC MODEL SD460

Size: 59 x	461	mm	
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100u A			€3.45
200u A		**	£3.40
500u A			£3.35
50-0-50u A			£3.45
100-0-100	uΑ		£3.40
1mA			£3.30
5mA			£3.30
			£3.30
			£3.30
			£3.30
500mA	••	••	£3.30

CLEAR PLASTIC MODEL MR 65P Size: 86 x 78mm

00uA £3.85 00uA £3.80 00uA £3.75 00-50uA £3.75			
100uA	iQuA		(
100uA	00u A	£3.85	1
100-500 A	:00uA	£3.80	- Armin
00.0-100uA	00u A	£3.75	A W
00.0-100uA	0-0-50u A	£3.85	7. L
ImA		£3.80	
ImA	00-0-500u A	£3.70	
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50mA £3.70 15V AC £3.80 100mA £3.70 50V AC £3.80 500mA £3.70 150V AC £3.80 500mA £3.70 150V AC £3.80 5A DC £3.70 500V AC £3.80 15A DC £3.70 500V AC £3.70 15A DC £3.70 VU Meter £4.70 15A DC £3.70 VU Meter £3.70 20A AC £3.70 50 AC £3.70 20A AC £3.70 20A AC £3.70 10V DC £3.70 30A AC £3.70 20V DC £3.70 50mA AC £3.70 20V DC £3.70 100mA AC £3.70 20V DC £3.70 200mA AC £3.70 20V DC £3.70 200mA AC £3.70			200V DC £3.70
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SV DC . £3.70 20A AC . £3.70 10V DC . £3.70 30A AC . £3.70 15V DC . £3.70 50mA AC . £3.70 20V DC . £3.70 100mA AC . £3.70 50V DC . £3.70 100mA AC . £3.70			404 40 4 60 70
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50V DC £3.70 200mA AC £3.70		43.70	
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500u A	£4.20
50-0-50u A	£4.45
100-0-100u A	£4.40
1mA	£4.20
1A DC ,,	£4.20
5A DC	£4.20
20V DC	£4.20
50V DC	£4.20
300V DC	£4.20
300V AC	£4.30
VU Meter	£4.70
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CLEAR PLASTIC MODEL MR 52P Size: 60 x 60mm

Size: 60 x 60mm		
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100uA	£3.50	(
500uA	£3.35	1 1 1
50-0-50uA	.€3.50	1
100-0-100u A	£3.45	WA W
1mA	£3.30	1
5mA	£3.30	
10mA	€3.30	
50m A	£3.30	
100mA	£3.30	
500mA	£3.30	Calling and an annual
1A DC	£3.30	
5A DC	€3.30	S Meter 1mA £3.
10V DC	£3.30	VU Meter £3.
20V DC	€3.30	1A AC * £3.
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300V DC	£3.30	10A AC * £3.
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300V	AC	::	::	£3.40				::		· £3.30	
BAK	ELI	TE	М	ODEL	MR	65	Siz	te: 8	10 x	80mm	
25uA				£5.25		6				-	

BAK	ELI	TE	MIU	UEL	MR	65	Size	: 81) x	80r	n
25uA				€5.25		4	COR				
50u A	٠			£4.00					-		۹
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500-	0-50	Du A		€3.60							
1mA				£3.60							
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10m	Α			£3.60		-			300		ļ
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100n	nΑ			£3.60		0V A			••	* £	
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1A D	C			£3.60		50V				°£	
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3.60	300V DC	 €3.60
	30V AC .	° £3.60
3.60		
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3.60	150V AC .	 ° £3.60
3.60	300V AC .	 * £3.60
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3 80	VU Meter.	 £4.10
3.60	1A AC .	 ° £3.60
3.60	5A AC .	 * £3.80
3.80	10A AC .	 ° £3.60
3.60	20A AC .	 * £3.60
3.60	30A AC .	 * £3.60
3.60	50A AC	 * £3.60

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500uA	£3.35	1 marian
50-0-50u A	€3.45	A
100-0-100uA	£3.40	0 0
1mA	£3.30	Children Company
5mA	£3.30	
10mA	£3.30	
50mA	£3.30	10V DC £3.30
100mA	£3.30	20V DC £3.30
500mA	£3.30	50V DC £3.30
1A DC	£3.30	300V DC £3.30
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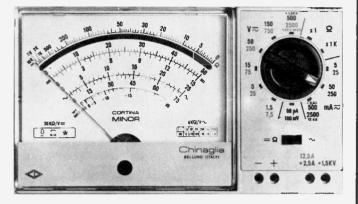


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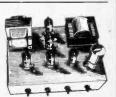
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149	60	3 12	9.9 × 7.7 × 8.6	3.98	45
150	100	5 8	9.9 × 8.9 × 8.6	4.45	45
151	200	8 0	12·1 × 9·3 × 10·2	7.39	53
152	250	13 12	12-1 × 11-8 × 10-2	8.93	73
153	350	15 0	14·0 × 10·8 × 11·8	10.80	73
154	500	19 8	14·0 × 13·4 × 11·8	12.41	91
155	750	29 0	17-2 × 14-0 × 14-0	18-65	
156	1000	38 0	17·2 × 16·6 × 14·0	26.50	
157	1500	46 0	21.6 × 13.4 × 18.1	30.23	
158	2000	60 0	21.6 × 15.3 × 18.1	33.70	
		ALITO	TRANSCORMERC		

				AU.	TO T	RAN	SFORE	1ERS		
Ref.	ŶA .	We	ight	S	ize cm		Auto	Tabs	P	& P
No.	(Watt	s) 1b	οz						- 6	D
113	20	1	٥.	5·8 ×	5-1 ×	4.5	0-115-21	0-240	1.52	30
64	75	2	4	7.0 ×	6.7 ×	6-1	0-115-21	0-240	2-64	38
4	150	3	4	8.9 ×	7-7 ×	7.7	0-115-20	0-220-240	3.75	45
66	300	6	4		9.6 ×		111	11	5-29	53
67	500	12	8	12·1 ×	11-2 ×	10.2	11	11	8.02	67
84	1000	19	8	14.0 x	13-4 x	14:3	in		12-44	91
93	1500	30	4	14.0 x	15-9 x	14-3	11.		16-65	
95	2000	32	0	17-2 ×	16.6 ×	14.0	11	11	22-00	
73	3000	40	ŏ	21.6 ×			***	**	31.90	

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ht Size cm. Secondary Winding:

4 (8 × 2 · 9 × 3 · 5 · 0 · 12 v at 0 · 25 A × 2

6 (1 × 5 · 8 × 4 · 8 · 0 · 12 v at 0 · 25 A × 2

7 (0 × 6 · 4 × 6 · 1 · 0 · 12 v at 0 · 5 A × 2

7 (0 × 6 · 4 × 6 · 1 · 0 · 12 v at 1 A × 2

8 · 3 × 7 · 7 × 7 · 0 · 0 · 12 v at 2 A × 2

8 · 9 × 8 · 0 × 7 · 7 · 0 · 12 v at 3 A × 2

9 · 9 × 8 · 0 × 7 · 7 · 0 · 12 v at 3 A × 2

9 · 9 × 9 · 6 × 8 · 6 · 0 · 12 v at 5 A × 2

12 · 1 × 9 · 9 × 10 · 2 · 0 · 12 v at 5 A × 2

12 · 1 × 9 · 9 × 10 · 2 · 0 · 12 v at 5 A × 2

14 · 0 × 9 · 6 × 11 · 8 · 0 · 12 v at 16 A × 2

14 · 0 × 9 · 6 × 11 · 8 · 0 · 12 v at 16 A × 2

17 · 2 × 15 · 3 × 14 · 0 · 12 v at 15 A × 2

17 · 2 × 15 · 3 × 14 · 0 · 12 v at 16 A × 2

ht Size cm. Secondary Tops

9 38 38 45 53 53 60 67 73 0.5 1.0 2.0 3.0 4.0 5.0 6.0 8.0 1.65 2.18 3.18 4.12 4.67 5.83 6.51 9.00 8.97 20 21 51 117 88 89 50 VÖLT KANGE

Weight Size cm. & P Ref Amps. Weight of the control 0·5 1·0 2·0 3·0 4·0 6·0 8·0 10·0 2·35 3·08 4·26 5·28 6·91 930 38 45 53 67 67 85 102 103 104 105 106 107 118

Amps. Weig 1b oz 0·5. 2 4 1·0 3 4 2·0 6 4 3·0 8 12 4·0 13 12 5·0 12 00 6·0 15 8 8·0 25 00 10·0 25 0 12·0 29 00 Weight Ref. No. 124 126 127 125 123 40 120 121 122 189 P 7.0 × 6.7 × 6.1 0.24-30-40-48-60V 8.9 × 7.7 × 7.7 9.9 × 9.6 × 8.6 12.1 × 9.9 × 10.2 12.1 × 11.8 × 10.2 14.0 × 10.2 × 11.8 14.0 × 12.1 × 11.8 14.0 × 12.1 × 11.8 17.2 × 12.7 × 14.0 17.2 × 12.7 × 14.0 17.2 × 14.0 × 14.0 738 38 45 60 67 73 85 2·12 3·10 4·62 6·84 7·96 8·87 10·27 13·64 15·93 18·16

MINIATURE TRANSFORMERS WITH SCREENS
mA Weight Size cm. Valts
lb oz L

2-8 × 2-6 × 2-0 6-1 × 5-8 × 4-8 3-9 × 2-9 × 3-5 6-1 × 5-4 × 4-8 7-0 × 6-4 × 6-1 4-8 × 2-9 × 3-5 6-1 × 5-8 × 4-8 7-0 × 6-1 × 6-1 8-3 × 7-7 × 7-0 8-9 × 7-7 × 7-7 No. 238 212 13 235 207 208 236 214 221 206 2 4 200 |A, |A |00 3-0-3 9-0-9 0-5, 0-6 9-0-9 0-8-9, 0-8-9 0-8-9, 0-8-9 0-15, 0-15 0-20, 0-20 20-12-0-12-20 0-15-27, 0-15-27 *Carriage 100 330, 330 500, 500 1A, 1A 200,200 300, 300 700 (d.c.) 1A, 1A 500,500 1 00 1 12 4 1 4 1 8 2 12 2 4 3 4

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MARRIOT XRPS/17 ± Track High £2-50; XRPS/18 ± Track Med. £3-50; XRPS/36 ± Track Med. £5-00; XRPS/63 ± Track High £1-75; Erase Heads for XRPS/17/18/36 (XES11) 75p; 13X12E 343

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ACY17	0 - 40	BY127	0 · 12	ZTX300	0 · 13
ACY39	0 - 78	BYZ13	0 · 42	ZTX302	0 - 18
AD149	0.50	C1060	0 - 54	ZTX500	0 - 13
AD161	0 - 44	GET111	0.72	2N697	0 · 16
AD162	0 - 44	GET115	0.90	2N706	0 12
AF117	0 - 24	GET880	0.60	2N930	0 - 18
AF118	0.57	LM309K	2.00	2N987	0 - 42
AF139	0 · 41	MAT 121	0-25	2N1132	0 - 24
AF186	0 · 48	MJE340	0 - 47	2N1304	0 - 28
AF239	0 - 44	MJE520	0.63	2N1613	0 - 21
ASY27	0.33	MJE3055	0.77	2N1671 2N2147	1·20 0·78
BA115	010	MJE 2955	1 - 27		0.78
BAX13	0 · 05	MPF105	0.36	2N2160 2N2926	0 - 78
BC107	0 - 14	NKT404	0.66		0.12
BC108	0 - 13	OA5	0.72	2N3053 2N3054	0.48
BC109	0 - 14	OA81	0 - 18	2N3054 2N3055	0 - 45
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BC113	0 - 15	OA202 OC28	0.66	2N3442	1-39
BC147	0 - 10	OC35	0.55	2N3525	0.91
BC148	0.08	OC36	0.60	2N3614	40.65
BC149	0 - 10	OC44	0 - 20	2N3702	0.11
BC169C	0-15	OC45	0 - 20	2N3714	1 - 41
BC182 BCY32	0 · 12 0 · 85	OC71	0.18	2N3771	1.77
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BCY70	0 - 18	OC81	0 - 29	2N3819	0 - 38
BCY71	0 - 22	OC83	0 - 27	2N3886	0.72
BCY72	0 - 12	OC140	1 - 14	2N3903	0 - 15
BD124	0.65	OC170	0 - 30	2N 4002	0 - 14
BD131	0 · 42	OC200	0 - 54	2N4126	0 - 15
BF115	0 - 20	OC202	0.90	2N4871	0.34
BF180	0 - 36	OCP71	1 · 20	2N5457	0.30
BF194	0 · 10	ORP12	0-60	25303	0.60
BFX13	0 · 26	ORP60	0.55	40550	0.54
BFX34	0.70	P346A	0 · 18	40361	0 - 45
BFX88	0 - 24	T1L209	0 - 20	40362	0 - 40
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TO5 1 Amp CRS1/05AF CRS1/10AF CRS1/20AF CRS1/20AF CRS1/60AF TO48 3 Amp CRS3/025AF CRS3/10AF CRS3/20AF CRS3/40AF CRS3/60AF TO48 7 Amp CRS7/400 CRS7/400	£ P 0 · 42 0 · 48 0 · 52 0 · 60 0 · 78 0 · 48 0 · 54 0 · 65 0 · 80	6 Amp SC40A SC40B SC40D SC40E 10 Amp SC45A SC45A SC45B SC45E 15 Amp SC50B SC50B	0-88 0-97 1-20 1-50 1-12 1-50 1-65 1-46 1-57 1-80 2-00
CRS/600 TO48 16 Amp CRS16/100 CRS16/200 CRS16/400	0·78 0·85 0·96	SC50E Aleo 40430 40669 40486	0·85 0·90 0·85

BRIDGE SILICON RECTIFIERS

} Amp	g 3	1	£ p	1 1 18	П.,
B025/025	£ p 0⋅14	B2/20	0 - 40	6 Amp	
B025/05	0.16	B2/40	0.44	B6/05	0 - 50
1 Amp	•	B2/60	0 - 45	B6/10	0 - 58
B1/05	0 - 20	B2/100	0.55	B6/20	0.68
B1/10	0.21	4 Amp		B6/40	0 · 75
B1/20	0 - 24	B4/05	0 · 45	B6/60	0 · 87
B1/60	0.25	B4/10	0 - 48	1 Amp Tu	bular
B1/100	0.30	B4/20	0.54	W005	0 · 27
2 Amp		B4 40	0.60	W01	0 - 29
B2/05	0 - 30	B4/60	0.70	W02	0.30
B2/10	0 - 35	B4/80	0.90	W06	0 - 33

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U4317 20Ω/V with case 16-50 U4341 33kΩ/V plus transistor steel

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U322 20K1/V plus 1kHz OSC, with case
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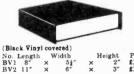
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BA3	4"	×	21"	×	14"	421
BA4	5±"	×	4"	×	11"	501
BA5	4"	×	24"	×	2"	421
BA6	3"	×	2*	×	1"	341
BA7	7*	×	ō*	×	21"	701
BA8	8"	×	6"	×	3"	901
BA9	6"	×	4"	×	2"	581
P. &	P. 10p	on e	sch box			

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ANTEX BIT	S and ELEMEN	ITS
Bits No.		
102 For model C	N240 32	421
104 For model C	N240 A"	421
1100 For model Co	CN240 🕺	421
1101 For model Co	UN240 1"	421
1102 For model Co	CN240 1"	421
1020 For model G	240 32	421
1021 For model G	240 #"	421
1022 For model G	240 %"	421
50 For model X	25 3 "	481
51 For model X	25 1"	48
52 For model X	25 Å″	481
ELEMENTS		
ECN 240 £1-30	ECCN 240 £1-32	
EG 240 £1-07	EX 25 £1-16	

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VAT included in all prices. Please add 10p P. & P. (U.K. only). Overseas ordersplease add extra for postage.

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BARG	MINS
Pack No. Qty.	Description Price
C1 200	Resistors mixed values approx. count by weight 0.55
C2 150	Capacitors mixed values approx. count by weight 0.55
C3 50	Precision Resistors 0.1%, 0.01% mixed values 0.55
C4 75	th W Resistors mixed preferred values 0.55
C5 5	Pieces assorted Ferrite Rods 0.55
C6 2	Tuning Gangs, MW/LW VHF 0.55
C7 1	Pack Wire 50 metres assorted colours 0.55
CB 10	Reed Switches 0.55
C9 3	Micro Switches 0.55
C10 15	Assorted Pots & Pre-Sets 0.55
C11 5	$\begin{array}{cccccccccccccccccccccccccccccccccccc$
C12 30	Paper Condensers preferred types mixed values 0.55
C13 20	Electrolytics Trans. types 0.55
C14 1	Pack assorted Hardware— Nuts/Bolts, Grommets, etc. 0-55
C15 5	Mains Slide Switches 0.55
C16 20	Assorted Tag Strips & Panels 0-55
C17 10	Assorted Control Knobs 0.55
C18 4	Rotary Wave Change Switches 0.55
C19 2	Relays 6-24V Operating 0.55
C20 1	Pack Sheets of Copper Laminate approx. 20 sq. ins. 0.55

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	Ref. 32A. Stylus Balance £1-37
	Ref. J. Tape Head Cleaning Kit 62p
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	Ref. 45. Auto Changer Groove Cleaner £1-08

Ÿ		OO AILD OOOILETO	
	PLUGS	,	
- 3	P8 1	D.I.N. 2 Pin (Speaker)	0.11
	P8 2	D.I.N. 3 Pin	0.12
- 1	PB 3	D.I.N. 4 Pin	0.15
- 3	P8 4	D.I.N. 5 Pin 180°	0.18
	PS 5	D.1,N. 5 Pin 240°	0.16
	PS 6	D.I.N. 6 Pin	0.17
	PS 7	D.I.N. 7 Pin	0-18
- 1	P8 8	Jack 2.5mm Screened	0.18
-	PS 9	Jack 3 5mm Plastic	0.12
s	PS 10	Jack 3.5mm Screened	0-18
9	PS 11	Jack 1" Plastic	0-15
	PS 12	Jack !" Screened	0-22
2p	PS 13	Jack Stereo Screened	0.36
2p		Phono	0.10
2p		Car Aerial	0-22
	PS 16	Co-Axial	0-15
2p	THETH	SOCKETS	
2р	PS 21		0.14
2p	P8 22		0.20
2p		D.I.N. 5 Pin 180°	0.20
2p		D.I.N. 5 Pin 240°	0-20
- 1	PS 25		0.16
8p	PS 96	Jack 3-5mm Plastic	0.16

PS 27 Jack 1" Plastic P8 28 Jack 1" Screened PS 29 Jack Steren Plastic Jack Stereo Screened PS 30 Phono Screened PS 32 Car Aerial

0.30

0.80

0-38

0.18

0.22

0.22

PS 33 Co-Axial

SOCKE	ETS	
PS 35	D.I.N. 2 Pin (Speaker)	0.08
PS 36	D.I.N. 3 Pin	0.11
PS 37	D.I.N. 5 Pin 180°	0.11
PS 38	D.I.N. 5 Pin 240°	0-11
PS 39	Jack 2-5mm Switched	0.12
PS 40	Jack 3 5mm Switched	0.12
P8 41	Jack ! Switched	0.20
P8 42	Jack Stereo Switched	0.30
PS 43	Phono Single	0.08
PS 44	Phono Double	0.10
PS 46	Co-Axial Surface	0.10
PS 47	Co-Axial Flush	0.20

LEADS

LS	1	Speaker	Lead	1 2	pin	D.	I.N.	pl	ug	te
		open o	enda	ap	prox	3	met	res		
		(coded	l)						0-	20

C	ΔB	LES	
CP	1	Single Lapped Screen	0.0
CP	2	Twin Common Screen	0.1
CP	3	Stereo Screened	0.15
CP	4	Four Core Common Screen	0.23
PC	5	Four Core Individually Screened	10.3
CP	6	Microphone Fully Braided Cabl	e 0·1
CP	7	Three Core Mains Cable	0.0
CP	8	Twin Oval Mains Cable	0.0
CP	9	Speaker Cable	0.0
CP	10	Low Loss Co-Axial	0.1
2	ΔR	RON	

	RBON	
Log at	nd Lin	
4.7K, 1M, 23	10K, 22K, 47K, 100K, 220K,	470K
VC 1	Single Less Switch	0.15
VC 2	Single D.P. Switch	0.28
VC 3	Tandem Less Switch	0-46
VC 4	1K Lin Less Switch	0.15
VC 5	100 K anti-Log	0.46
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0.1 watt 0.06 each 100, 220, 470, 1K, 2·2K, 4·7K, 10K, 22K, 47K, 100K, 220K, 470K, 1M, 2M, 4·7M

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ype	Amps	Price	P. & P.
T50/i		£1.93	30p
T50/1		£2.42	85p
T50/2		£3.30	40p

CARTRIDGES

4.000	
ACOS	
GP91-18C 200mV at 1-2cm/sec	£1.85
GP93-1 280mV at 1cm/sec	\$1-85
GP96-1 100mV at 1cm/sec	£2-80
J-2005 Crystal/Hi Output	£1-05
J-2010C Crystal/Hi Output Compa	tible £1-20
J-20068 Stereo/Hi Output	21-75
J-2105 Ceramic/Med Output	£1-95
J-2203 Magnetic 5mV/5cm/sec,	including
stylus	24.95
J-22038 Replacement stylus for ab	ove \$3.00
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4mV/5cm/sec	£8-30

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The E12 Range of Carbon Film Resistors,

1 watt available in PAKS of 50 pleces,
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R1 50 Mixed 100 ohms-820 ohms

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850 Mixed 100 -82 kΩ

850 Mixed 100 Ω-82 kΩ

850 Mixed 100 Ω-82 kΩ

850 Mixed 100 Ω-100 Ω

850 Mixed 100 Ω

850 Mixed 10

BI-PAK SUPERIOR QUALITY LOW - NOISE CASSETTES

C60, 36p; C90, 48p; C120, 60p.

-the lowest prices!

BI-PAK QUALITY COMES TO AUDIO!

AL10/AL20/AL30 AUDIO AMPLIFIÉR MODULES



The AL10, AL20 and AL30 units are similar in their appearance and in their general specification. However, careful selection of the plastic power devices has resulted in a range of output powers from 3 to 10 watts R.M.S.

The versatility of their design makes them ideal for use in record players, taperecorders, stereo amplifiers and cassette and cartridge tape players in the car and at home.

Conditions	Performance
	Lettokmance
Po = 3 WATTS f = 1KHz	0.25%
	8-16Ω
f = 1KHz	100 k Ω
Po = 2 WATTS	50 Hz-25KHz
$V_8 = 25V$. $R1 = 8\Omega$ $f = 1KHz$	75mV. RM8
_	3" × 21" = 1"
	Po = 3 WATT8 f = 1KHz f = 1KHz

The above table relates to the AL10, AL20 and AL30 modules. The following table outlines the differences in their working conditions.

Parameter	AL10	AL20	AL80
aximum Supply Voltage	25	30	30
ower out for 2% T.H.D. (RL = 80 f = 1KHz)	3 watts RMS Min.	5 watts RMS Min.	10 watts RMS Min.

AUDIO AMPLIFIER MODULES

AL 10. 3 watts AL 20. 5 watts AL 30. 10 watts

POWER SUPPLIES

POWER SUPPLIES
PS 12. (Use with AL10, AL20, AL30) 95p
SPM 80. (Use with AL60)
\$23.25
FRONT PANELS FP 12 with Knobs
\$1.00

PRE-AMPLIFIERS

PA 12. (Use with AL10, AL20 44-35 and AL30) PA 100. (Use with AL60)

TRANSFORMERS

T461 (Use with AL10) #1-60 P & P 15p T461 (Use with AL10) £1.60 F & F 10p T538 (Use with AL20, AL30) £2.30 P & P 15p

BMT80 (Use with AL60) 22.75 P & P 25p

PA12 PRE-AMPLIFIER SPECIFICATION

The PA12 pre-amplifier has been designed to match into Frequency response—20Hz-50KHz(-3dD) most budget stereo systems. It is compatible with the AL 10, AL 20 and AL 30 audio power amplifiers and it can be supplied from their associated power supplies. There are two stereo inputs, one has been designed for use with *Ceramic cartridges while the auxiliary input will suit most †Magnetic cartridges. Full details are given in the specification table. The four controls are, from left to right: Volume and on/off switch, balance, bass and treble. Size 152mm × 84mm × 35mm.

± 12dB at 60Hz
Treble control— Treble control— ± 14dB at 14KHz *Input 1. Impedance 1 Meg. ohm Sensitivity 300mV †Input 2. Impedance 30 K ohms

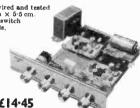
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ALL PRICES INCLUDE V.A.T.

The STEREO 20

The "Stereo 20" amplifier is mounted, ready wired and tested on a one-piece chassis measuring 20 on × 14 cm × 5.5 cm.
This compact unit comes complete with on/off awitch volume control, balance, bass and treble controls,
Transformer, Power supply and Power amps.
Attractively printed front panel and matching control knobs. The "Stereo 20" has been designed to fit into most turntable plints without interfering with the mechanism or, alternatively, into a separate cabinet.
Output power 20 w peak. Input 1 (Cer.) 300mV into 1M. Freq. res. 25th-25tkHz.
Input 2 (Aux.) 4mV into 30K. Harmonic distortion. Bass control ± 12dB at 60Hz typically 0.25% at 1 watt. Treble con.
± 14dB at 14kHz.



TC20 TEAK VENEERED CABINET

For Stereo 20 (front board undrilled) Size 104" × 84" × 3", \$3.95 plus 30p postage.

SHP80 STEREO HEADPHONES

4-16 ohms impedance. Frequency response 20 to 20,000Hz. Stereo/mono switch and volume controls, 24-95

NOW WE GIVE YOU 50w PEAK (25w R.M.S.) **PLUS THERMAL PROTECTION!** The NEW AL60 Hi-Fi Audio Amplifier FOR ONLY £3.95

- Max Heat Sink temp 90°C.
- Frequency Response 20Hz to 100KHz
- Distortion better than 0.1% at 1KHz
- Supply voltage 15-50 volts
- Latest Design Improvements
- Load 3, 4, 8 or 16 ohms Signal to noise ratio 80dB
- Overall size 63mm × 105mm

Especially designed to a strict specification. Only the finest components have been used and the latest solid state circuitry incorporated in this powerful little amplifier which should satisfy the most critical A.F.

STABILISED POWER **MODULE SPM80**

SPM80 is especially designed to power 2 of the AL60 Amplifiers, up to 15 watt (r.m.s.) per channel simultaneously. This module embodies the latest components and circuit techniques incorporating complete short circuit protection. With the addition of the Mains Transformer BMT80, the unit will provide outputs of up to 1-5 amps at 35 volts. Size: 63mm × 105mm × 30mm.

These units enable you to built Audio Systems of the highest quality at a hitherto unobtainable price. Also ideal for many other applications including:—Disco Systems, Public Address, Intercom Units, etc. Handbook available 10p PRICE £3-25

TRANSFORMER BMT80 £2:15 p. & p. 28p

STEREO PRE-AMPLIFIER TYPE PA100

Built to a specification and NOT a price, and yet still the greatest value on the market, the PA100 stereo pre-amplifier has been conceived from the latest circuit techniques. Designed for use with the AL60 power amplifier system, this quality made unit incorporates no less than eight sliicon planar transistors, two of these are specially selected low noise NPN devices for use in the input stages.

Three awitched stereo inputs, and rumble and scratch filters are features of the PA100, which also has a STEREO/MONO switch, volume, balance and continuously variable bass and treble controls.

SPECIFICATION

SPECIFICATION	Frequency Response	20Hz - 20KHz ± 1dB
Harmonic Distortion	525 mV into 50KΩ	3.25 mV into 50KΩ
2. Radio, Tuner	75 mV into 50KΩ	3 mV into 50KΩ
All input voltages are for an output of 250mV. Tape and P.U. inputs equalised to RIAA curve within ± 1dB. Irom 20Hz to 20KHz.		
Bass Control	15dB at 20KHz	15dB at 20KHz
15dB at 20KHz	15dB at 20KHz	
15dB at 20KHz	15dB at 20KHz	
15dB at 20KHz	15dB at 20KHz	
15dB at 20KHz	15dB at 20KHz	
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15dB at 20KHz	15dB at 20KHz	
15dB at 20KHz	15dB at 20KHz	15dB at 20KHz
15dB at 20KHz	15dB at 20KHz	15dB at 20KHz
15dB at 20KHz	15dB at Treble Control	
Filters: Rumble (High Pass)
Scratch (Low Pass)

Scratch (Lo Signal/Noise Ratio Input overload Supply Dimensions

100Hz

8KHz better than -65dB + 26dB + 35 volts at 20mA 292mm × 82mm × 35mm

ONLY £13.15

MK 60 AUDIO KIT
Comprising: 2 × AL80, 1 × 8PM80, 1 × BTM80, 1 × PA 100, 1 front panel, 1 kit of parts to include on-off switch, neon indicator, stereo headphone sockets plus instruction booklets. Complete Price: 228-75 plus 30p postage.

TEAK 60 AUDIO KIT

Comprising: Teak veneered cabinet size $16\frac{\pi}{4} \times 11\frac{\pi}{4} \times 3\frac{\pi}{4}$, other parts include aluminium chassis, heatsink and front panel bracket, plus back panel and appropriate sockets, etc. Kit price: 29.95 plus 30p postage.

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Varnish Impregnated Size 45mm × 36mm × 31mm

PRI 240V 100m A Sec 6.0.6 100mA Sec 9.0.9 100mA Sec 12.0.12 100mA Sec 20.0.20 100mA £1.23 10p P, & P.

CRESCENT BUBBLE LIGHT SHOW This budget system compares very favourably with more sophisticated and higher priced models

Specification:
Projector—150W convection
cooled. At 30ft the projected
image is 16ft.
Motor—1 rev. per 2 min.

Motor—1 rev. per 2 min. Liquid Wheel—6in diameter

Liquid Wheel—6in diameter multi colour.

The motor is fitted to the projector and can only be purchased as a

single unit.

The liquid wheel is our standard model and may be purchased separately.

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7in ×4in LOUDSPEAKER A top quality speaker ideal where small size is import-



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PPI Switched 3-6-71-9V 400M/A Transistor and
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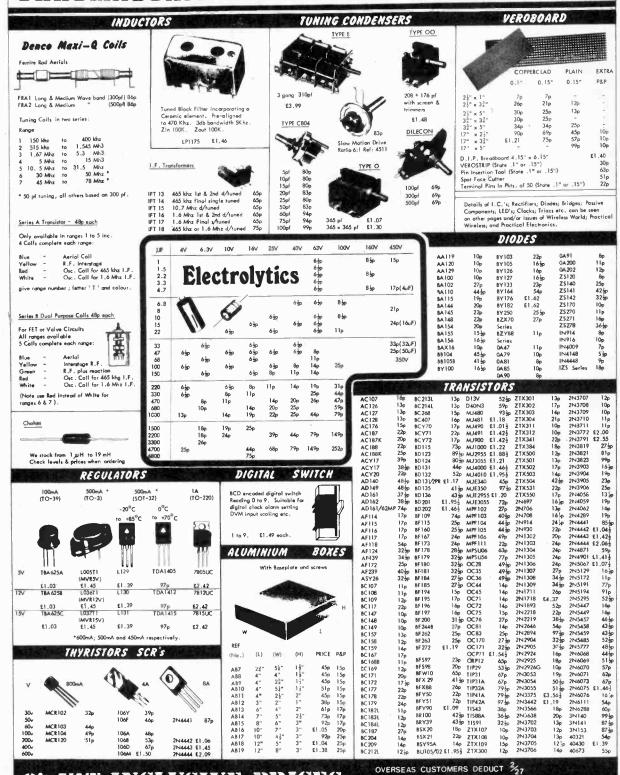
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	4	1	£р	1	£p	Ĭ	£р		£p 0-10
1N21	£p 0·17	FAZ11	1.15	BY213	0.25	OAZ205	0.45	Z8170	0.10
IN23	0.85	AFZ12	2.00	BYZ10	0.45	OAZ206	0.45	Z8271	0.18
IN85	0.88	ASY26	0.25	BYZ11	0.40	OAZ207	0.45	ZT21 ZT43	0.25
1N253	0.50	ASY27 ASY28	0.30	BYZ12	0.40	OAZ208 OAZ209	0·40 0·40	ZTX107	0·25 0·12
1N256	0.50	A8 Y 29	0.25 0.20	BYZ13	0.35	OAZ210	0.40	ZTX108	0-10
IN646	0.16	A8V36	0.25	BYZ15	1.25	OAZ211	0.40	ZTX300	0-14
1N725A 1N914	0.20	ASY36 ASY50	0.20	BYZ16	0-60	OAZ222	0.45	ZTX304	0.24
1N4007	0.12	ASY51	0.40	BZY88	0.10	OAZ223	0.45	ZTX 500	0-15
18113	0.25	ASY53	0.20	C111	0.55	OAZ224	0.45	ZTX 503	0.16
18131	0-18	ASYōō	0.20	CR81/05	0.30	OAZ241	0.25	ZTX531	0.25
18202	0.28	A8Y62	0.25	CR81/40 CS4B	0·45 1·90	OAZ242	0.15	THEFT	A STATE OF
2G371	0.40	ASY66 ASZ21	0.33 1.00	C810B	3-50	OAZ244 OAZ246	0-25 0-15	CIRCUIT	LIED
2G381	0.22	ASZ23	0.75	DD000	0-15	OAZ290	0-28		
20414	0.30	AU101	1.50	DD003	0.15	OC16	1.00	7400	0.80
20417	0.25	AU101 AUY10	1.00	DD006	0.25	OC16T	1.00	7401	0-20
2N404 2N697	0-22 0-15	RC107	0.12	DD007	0.40	OC19	0-50	7402	0-20 0-20
2N698	0.80	BC108 BC109 BC113	0.12	DD008	0.88	OC20	2.00	7404	0.20
2N706	0.10	BC109	0.12	GD3	0.88	OC22	1.00	7405	0.20
2N706A	0.12	BC113	0-16	GD4	0-10	OC23	1.25	7406	0-40
2N708	0.15	BC115 BC116	0-20	GD5 GD8	0-38 0-25	0C24	1.10	7407	0-40
2N709	0.40	BC116	0-20	GD12	0.10	OC25 OC26	0-40	7408	0.25
2N1091	0.55	BC116A BC118	0-23 0-20	GET102	0.50	OC28	0-40	7409	0-88
2N1131	0.25	BC121	0-20	GET103	0-40	0029	0-65	7410	0-20
2N1132	0.25	BC121	0-20	GET113	0.85	OC30	0-40	7411	0.28
2N1302	0.18	BC122 BC125	0.68	GET114	0.80	OC35	0.55	7412	0.80
2N1303 2N1304	0·18 0·22	BC126 BC140	0.65	GET115	0.75	OC35 OC36	0.65	7413 7416	0.80
2N1304	0.22	BC140	0.55	GET116	0-85	OC41	0-85	7417	0.80
2N1306	0.28	BC147	0.12	GET120 GET872	0-50 0-30	OC42 OC43	0-40 0-70 0-18	7420	0.20
2N1307	0.28	BC148	0.10	GET875	0.40	OC43	0-70	7422	0.28
2N1308	0.28	BC149 BC157	0.12	GET880	0-55	OC44M	0.17	7423	0-40
2N2147	0.75	BC154	0·14 0·12	GET881	0.25	0045	0-18	7425	0.87
2N2148	0.60	BC158 BC160	0.63	GET882	0-85	OC45 OC45M	0.18	7427	0-87
2N2160	1.00	BC169	0.14	GET885 GEX44	0-40	OC46	0.27	7428 7430	0-48
2N2218 2N2219	0.28	BC169 BCY31 BCY32 BCY33	0.45	GEX44	0.08	OC57	0-80	7432	0-20 0-87
2N2369A	0.16	BCY32	1.20	GEX45/1	0.45	OC58	0.80	7433	0-48
3N2444	1.99	BCY33	0.88	GEX941	0-45	OC59	0-60	7437	0.43
2N2613	0.28	BCY34	0.45	GJ3M	0.50	OC66	0-50	7438	0.48
2N2646	0.50	BCY34 BCY38 BCY39	0.55	GJ4M GJ5M	0.25	OC70 OC71	0-18 0-15	7440	0-20
2N2904	0.20	BCY39	1-00 0-80	GJ7M	0.50	0071	0.25	7441AN	0.85
2N2904A	0.25	BCY40 BCY42 BCY70	0.80	HG1005	0.50	OC72 OC73 OC74	0-50	7442	0-85
2N2906	0.20	BCV70	0.15	H8100A	0-20	OC74	0.80	7450 7451	0-20 0-20
2N2907 2N2924	0.18	BCY71 BCZ10	0.20	MAT100	0-20	OC75 OC76 OC77	0-80	7453	0.20
2N2925	0.15	BCZ10	0.60	MAT101	0.25	OC76	0.80	7454	0-20
2N2926	0.10	BCZ11	0.65	MAT120 MAT121	0.20	OC77	0.55	7460	0-20
2N3054	0-50	BD121	1.00	MJE520	0-25	OC78 OC79	0-25	7470	0.88
2N 3055	0-60	BD123	1.00 0.80	MJE2955		OC81	0-28	7472	0.88
2N3702	0.11	BD124 BDY11	1.45	MJE3055	0-75	OC81D	0-28	7473	0.44
2N 3705	0.15	BF115	0.22	MJE340	0.50	OC81M	0.20	7474 7475	0.48
2N3706	0-11	BF117	0.50	MPF102 MPF103	0-40	OC81DM	0.18	7475	0.45
2N3707 2N3709	0·13 0·10	BF167	0.25	MPF103	0-86	OC81Z	0.45	7480	0.80
2N3710	0-11	RF173	0.28	MPF104	0-85	OC82	0.28	7482	0.87
2N3711	0-11	BF181 BF184	0-35	MPF105	0.46	OC82D	0-25	7483	1.20
2N3819	0-85	BF184	0-22	NKT128 NKT129	0.45	OC83	0.25	7484	1.00
2N4289	0.20	BF183 BF194	0.22	NKT211	0.25	OC84 OC114	0.80	7486	0-50
2N5027	0.53	BF194 BF195	0·13 0·18	NKT213	0.25	OC122	1.00	7490	0.75
2N5088	0-33	RF196	0-15	NKT214 NKT216	0.24	OC123	1.10	7491A	1·10 0·75
28301	0.59	BF196 BF197	0.15	NKT216	0-40	OC139	0.40	7492 7493	0.75
26804 28501	1·15 0·75	BF861	0.25	NKT217 NKT218	0-45	OC140	0.65	7494	0-75 0-85
28703	1.00	BURNER	0-25	NKT218	1.13	OC141 OC169	0.80	7403	0-85
AA129	0-20	BFX12 BFX13	0-20	NKT219	0-88	OC169	0.20	7496	1.00
AAZ12	0-75 0-10	BFX13	0.25	NKT222	0-80 0-25	OC170	0.25	7497	4-82
AAZ13	0-10	BFX29 BFX30	0-28	NKT222 NKT224 NKT251	0.24	OC171 OC200	0.80	74100	2-16
AC107	0.85		0-28 0-98	NKT271	0.20	OC201		74107	0.51
AC126	0.25	BFX63 BFX84 BFX85 BFX86 BFX87	0.50	NKT271 NKT272 NKT273	0-20	OC202	0.80		0-57 0-86
▲ 0127	0.25	BFX84	0-25	NKT273	0.20	OC203 OC204	0-55	74111 74118	1.00
AC128 AC187	0.20	BFX85	0.28	NKT274 NKT275 NKT277	0-20	OC204	0.65	74119	1.92
AC188	0.20	BFX86	0.25	NKT275	0-25	OC205 OC206	1.00	74121	0-57
ACY17	0.35	BFX87	0.25	NKT277	0-20 0-25	OC206	1.10	74122	0.80
ACY18	0.27	DIAGO	0.22	NKT278 NKT301	0-85	OC207	1-00 0-20	74123	1-44
ACY19	0.27	BFY10 BFY11	1-00	NKT304	0.75	OC460 OC470	0-80	74141	1-00
ACY20	0.55	BFV17	9-50 0-40	NKT403	0-70	OCP71	1.00	74145	1.44 2.80
ACY21	0.22	BFY18	0.45	NKT403 NKT404	0.60	ORP12	0.55	74150 74151	1.15
ACY22	0.25	BFV19	0-55	I NKT678	0-80	ORP60	0-45	74154	2.80
ACY28	0.25	BFY24	0.45	NKT713	0.80	ORP61	0.48	74155	1.15
ACY39	0.65	BFY44	1.00	NKT773 NKT777	0.25	8X68	0-20	74156	1.15
ACY40	0.22	BFY50 BFY51	0.20 0.20	078B	0.38	8X 631 8X 635	0-45 0-55	74157	1.09
ACY41	0.22	BFY52	0.20	OA5	0.60	8X 640	0.75	74170	2.88
ACY44	0.82	BFV53	0.17	OA6	0.12	8X641	0.75	74174	1.80
AD140	0.50	BFY53 BFY64	0.45	OA47	0-08	8X642	0.60	74173	1.29
AD 149 AD 161	0.50	BFY90	0.75	OA70	0.10	8X644	0.85	74176	1-44
AD162	0.89	BSX27	0.50	OA71	0.20	SX 645	0.85	74190	2.80
AF106	0-80	BBX60	0.98	OA73	0·15 0·15	TIC44	0-29	74191	2.30
AF114	0.25	B8X76	0.18	OA74 OA79	0.10	V15/30P	0.75	74192	2.80
AF115	0.25	BSY26	0.17	OA81	0.10	V30/201P		74193	2.80
AF116	0.25	BSY27 BSY51	0.20 0.50	OA85	0.15	V60/201	0-50	74194	1-72
AF117	0.20	BSY95A	0.12	OA86	0.15	V60/201P	0.75		1.44
AF118 AF119 -	0.50	BSY95	0.12	OA90	0.07	X A 101	0.10	74195	
AF124	0.80	BT102/50	10 R	OA91	0-07 0-07	XA102	0-18	74196	1.58
AF125	0.30		0.75	OA95	0.07	XA151	0.15	74197	1.58
AF126	0-80	BTY42	0.92	OA200	0-08	XA152	0.15	74198	8-16
AF127	0-80 0-88	BTY79/1	00R 0-75	OA202 OA210	0.20	XA161	0.25	74199	2-88
AF139	0.88	BTY79/4		0A211	0-25	XA162	0.25	Plug in	ckot-
AF178	0-55	D1110/4	1.10	OAZ200	0-50	XB101	0-48	Plug in so —low pro	file
▲F179 ▲F180	0.65	BY100	0.15	OAZ201	0-45	XB102	0.80	14 pin DI	L
AF180 AF181	0.55	BY126	0-14	OAZ202	0-45	XB103	0-85		0.15
AF186	0.40	BY127	0-15	OAZ203	0-45	XB113	0-80	16 pin DI	L
AFY19	1.13	BY182	0.85		0.45	XB121	0-43		0-17
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PUSH BUTTON CAR RADIO KIT

The Tourist II



NO SOLDERING REQUIRED!

NOW BUILD YOUR OWN PUSH BUTTON CAR RADIO

Easy to assemble construction kit comprising fully completed and tested printed circuit board on which no soldering is required. All connections are simple push fit type making for easy assembly. Fine tuning push button mechanism is fully built and

Car Radio Kit £7.70 + 55p p.8 p

tested to mate with printed circuit board.

The Tourist | Kit For the experienced constructor | If you can solder on a printed circuit board you can build this model. Same technical specification as Tourist || Price £6.60 + 55p p & p.

Technical specification:

- (1) Output 4 watts R.M.S. output. For 12 volt operation on negative or positive earth.
- (2) Integrated circuit output stage, pre-built three stage IF Module.

Controls volume manual tuning and five push buttons for station selection, illuminated tuning scale covering full, medium and long wave bands. Size chassis 7" wide, 2" high and 4%" deep approx

Speaker including baffle and fixing strip £1.65+23p. p&p. Car Aeriał Recommended — fully retractable £1.37+20p. postage & packing



STEREOT QUA

QUALITY SOUND FOR LESS THAN £20.00

Stereo 21, easy to assemble audio system kit. No soldering required. The unit is finished in white P.V.C. and the acrylic top presents an unusually interesting variation on the modern deck plinth.

Includes:— BSR 3 speed dock, automatic, manual facilities together with ceramic cartridge.

Two speakers with cabinets,
Amplifier module. Ready built with control panel, speaker leads and full, easy to follow assembly instructions.
Specifications: For the technically minded:—

Input sensitivity 600mV. Aux. input sensitivity 120mV. Power output 2.7 watts per channel. Output impedance 8–15 ohms. Stereo headphone socket with automatic speaker cutout. Provision for auxiliary inputs – radio, tape, etc., and outputs for taping discs. Overall Dimensions. Speakers approx. $15\frac{y}{2} \times 8^{\prime\prime} \times 4^{\prime\prime}$. Complete deck and cover in closed position approx. $15\frac{y}{2} \times 12^{\prime\prime} \times 6^{\prime\prime}$. Complete $0 \times 10^{\prime\prime} \times 10^{\prime\prime} \times 10^{\prime\prime} \times 10^{\prime\prime} \times 10^{\prime\prime}$. Complete only £19-95 +£1.60 p & p. Extras if required. Optional Diamond Styli £1.37.

Complete only L 19:95 + £1.60 p & p. Extras if required. Optional Diamond Styli £1.37.

Specially selected pair of stereo headphones with individual level controls and padded earpieces to give optimum performance, £3.85.



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8TRACK HOME CARTRIDGE PLAYER *



Elegant self selector push button player for use with your stereo system. Compatible with Viscount III system, Unisound module and the Stereo 21. Technical specification Mains input, 240V. Output sensitivity 125mV Comparable unit sold eleswhere at £24.00 approx. Yours for only

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COMPLETE* STEREO SYSTEM



System 1. £51.00

40 Watt Amplifier. Viscount III - R102 now 20 watts per channel. System Lincludes

Viscount III amplifier - volume, bass, treble and balance controls, plus switches for mono/ stereo on/off function and bass and treble filters. Plus headphone socket.

Specification

20 watts per channel into 8 ohms. Total distortion @ 10W @ 1kHz 0·1%. P.U.1 (for ceramic cartridges) 150mV into 3 Meg. *P.U.2* (for magnetic cartridges) 4mV @ 1kHz into 47K. equalised within =1dB R.I.A.A. *Radio* 150mV into 220K. (Sensitivities given at full power). Tape out facilities: headphone socket, power out 250mW per channel. Tone controls and filter characteristics. Bass: +12dB to -17dB @ 60Hz. Bass filter: 6dB per octave cut. Treble control: treble + 12dB to -12dB @ 15kHz. Treble filter: 12dB per octave. Signal to noise ratio: (all controls at max.) – 58dB. Crosstalk better than 35dB on all inputs. Overload characteristics better than 26dB on all inputs. Size approx. $13\frac{3}{4}$ \times 9 \times $3\frac{3}{4}$

Garrard SP 25 Mk III deck with magnetic cartridge, de luxe plinth and hinged cover. Two Duo Type II matched speakers — Enclosure size approx. $17\frac{\pi}{2}$ × 10^2 × 8° in simulated teak. Drive unit 13° × 8° with parasitic tweeter. 10 watts handling.

Complete System £51.00

ystem 2. £.69·00

Garrard SP 25 Mk III deck (As System I)

Two Duo Type III matched speakers - Enclosure size approx. 27" x 13" x 11½" Finished in teak veneer. Drive units 13" x 8" bass driver, and two 3" (approx.) tweeters. 20 watts R.M.S., 8 ohms frequency range — 20 Hz to 18,000 Hz.

Complete System £69.00

PRICES: SYSTEM 1

Viscount III R102

amplifier

£24.20 + £1 p & p

2 Duo Type II speakers £14.00 + £2.20 p & p

Garrard SP 25 with Mag. cartridge

de luxe plinth and hinged cover

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Rass Unit

Flux density-100 K, speech coil-12 Cone, Triple laminated paper with P.V.C. surround. Mid Range Unit

Flux density-33K, speech coil-1" with

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950

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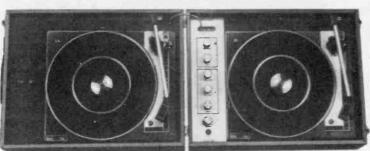
PORTABLE DISCO CONSOLE

INCORPORATES: Pre-Amp with full mixing facilities, including switched input for mic with volume control, switched input for auxiliary with volume control, bass and treble controls. volume control and blend control for turntables.

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TECHNICAL SPECIFICATION:

Pre-amp — Output - 200mV. Auxiliary inputs - 200mV and 750mV into 1 meg. 240 volt operation. Mic input - 6mV into 100K.

Turntables capacity - 7", 10" or 12" records.

Rumble, wow and flutter - Rumble - Better than -35dB. Wow - Better than 0.2%. Flutter - Better than 0.06% (Gaumont kalee meter).

Finish - Satin black mainplate with black turntable mat inlaid with brushed aluminium trim. Tonearm and controls in black and brushed aluminium.

Console size - Unit Closed - 17 3" × 13 3" × 8 3" (approx.) Unit Open - 35 3"× 13 3"× 4 3" (approx.)

This disco console is ideally matched for the Reliant IV and Disco 50 or any other quality amplifier.

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ISCO 5



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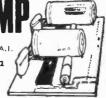
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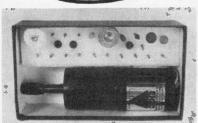
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10	_	_	_	_	8p	9p	8p	8p
22	_	_	8p	_	9p	8p	8p	10p
47	8p	_	9p	8p	8p	8p	10p	13p
100	Sp.	8p	8p	8p	9p	10p	12p	19p
220	8p	8p	1p	10p	10p	11p	17p	28p
470	9p	10p	10p	11p	13p	17p	24p	45p
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5 amp, 6, 8, 10, 12, 16, 18, 20, 24, 30, 36, 40, 48, 60 26-00

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5 amp, 6, 8, 10, 12, 16, 18, 20, 24, 30, 36, 40, 48, 60 26-00

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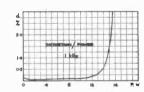
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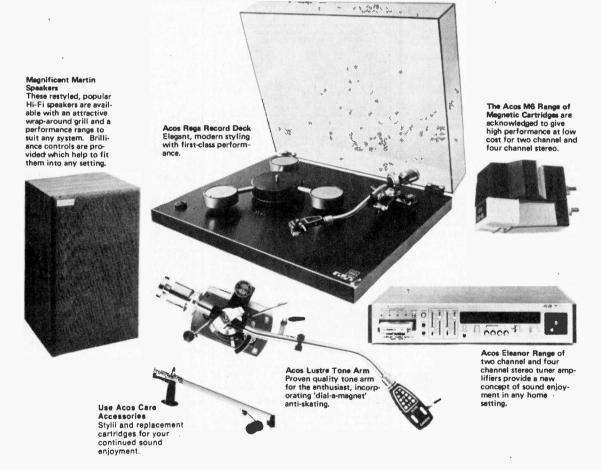
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THE TRIUMPHS OF RADIO ASTRONOMY

The charting of unimaginable depths of space has been assiduously carried out by radio astronomers in a number of countries since this science was formally established some 30 years ago. In Britain we can be especially proud of the achievements of our own radio astronomers, in particular those at Mullard Radio Astronomy Observatory, Cambridge, who have reached out to the furthermost parts of the Universe. And now has come international recognition and reward. Two scientists chiefly responsible for the significant discoveries at Cambridge have been awarded the 1974 Nobel Prize for Physics (see page 38).

The triumphs of radio astronomy are scintillating examples of pure science working hand in hand with technology in the quest for knowledge of the very Universe itself. Undoubtedly this young science does deserve more general

recognition.

Lacking any dramatic dynamic action and readily assimilatable evidence like, for example, that which accompanies each Space Shot, radio astronomy has remained a distant and little understood or appreciated science so far as the general public is concerned. Radio telescopes in the popular mind are parabolical structures that can be rotated and aimed at any desired part of the sky, as exemplified by the very familiar Jodrell Bank telescope. In reality, radio telescopes can take diverse forms, and as often as not consist of static arrays of aerials bearing no resemblance to the general conception of a "telescope".

The emissions received from the far-off radio sources are recorded and provide data for computers and mathematicians to digest. It is only after the scientists have performed their analysis that the results can be presented in more tangible form that will be meaningful to the non-expert. This is of course in contrast to optical astronomy. Armed with even a modest telescope, anyone can make observations from his back garden. The results obtained will differ, essentially, only in degree of magnitude of resolution and distance covered with those of the professional optical astronomers.

With radio astronomy, as we have indicated, the observer has to be an expert interpreter, while the apparatus employed will comprise complex aerial arrays and highly sensitive receiving equipment—often unique and specially developed by members of a radio astronomy observatory staff for particular kinds of observations. Thus at both levels, science and technology, radio astronomy is almost exclusively the domain of the professionals. And in addition, so far as the larger observatories are concerned, generous financial backing is essential, from government or industrial bodies.

This does not mean that the amateur is precluded from undertaking any investigations in radio emissions from extraterrestrial bodies. But it does mean that from practical considerations he is restricted to observations of the Solar System, our own Galaxy, and the more powerful sources of nearby galaxies. Here he will be following in the footsteps of the professionals, whose early work included extensive examination of these regions and, we should note, they themselves were following the pioneer investigations made by an American amateur named Grote Reber. It is indeed all too easy to assume that no significant discoveries are likely to reward amateur efforts today. But does one really know for sure?

F.E.B.

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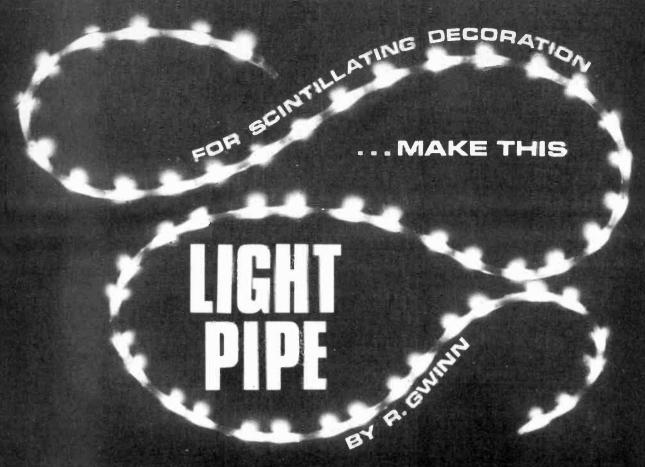
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OBILE light displays are all the rage nowadays with multi-acloured displays driver by amplifiers a common feature of parties and dances. The flexibility of electronics has not yet been exploited fully in this area but this article, describing a plastic pipe down which light can be made to seem to flow, shows the potential available to constructors with imagination.

The light pipe consists of a plastic tube 24ft long, containing 144 miniature light bulbs which are flashed in a sequence which makes barics of light

appear to move down the tube.

The bulbs are conrected in four series chains of 36 bulbs, and at any time two adjacent bulbs are or and wo are off. This pattern is repeated 36 times down the tube; and to give the effect of motion, the rear bulb of a pair is turned off and the bulb in front is turned on. This sequence is shown in Fig. 1.

SEQUENCING LOGIC

The generation of the sequence switching a performed by five TTL i.e.s. In fact a variable pulse generator clocks two flip-flops which count up to four, providing the four possible states. The two-bit binary is decoded into one-of-four by four NAND gates in a 7400.

Besides decoding, these gates also invert the outputs. This is convenient because he next operation requires on gating to turn on control channels for two consecutive states. This can be cone with NAND gates, as these effectively become now gates in negative lagic, which is the result of the one-of-four outputs being inverted.



Fig. 1. The first ten bulbs in the Light Pipe showing the four switching states, 0 being the off condition and 1 the on condition. The diagonal patter indicates the apparent direction of light motion

THE CIRCUIT

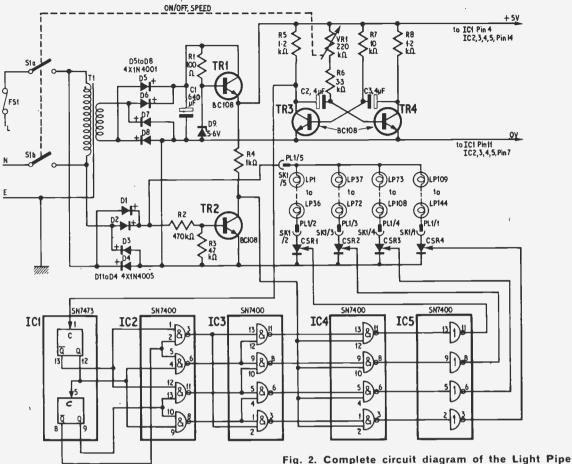
The circuit diagram of Fig. 2 indicates the basic simplicity of the light pipe. Mains power drives he logic via a power supply circuit and a variable pulse generator, whilst the strings of bulbs in the light pipe are switched by thyristors CSR1, 2, 3 and 4 inder control of the logic output from IC5.

INTERFERENCE SUPPRESSION

IC4 gates through the information to IC5 orly when the supply to the lamps, which it can be seen is pulsating d.c., is low. This is to reduce interference which could be caused if a large current was switched on very quickly by a thyristor.

As the hipe can be draped round lourspeakers and amplifiers, which if hadly screened can pick up an objectionable clicking, it is good practice no o generate interference in the first place rather than

try to eliminate it later.



IC5 inverts the signals and directly drives the gates of the thyristors.

The zero-crossing pulses for IC4 are produced by TR2, which is on until the supply to the lamps falls below a given level.

POWER SUPPLIES

The logic 5V is obtained from a 6V transformer, rectified to give 8V d.c. regulated at 5V. The lamps are driven from full-wave rectified mains, so that the thyristors are operational in both positive and negative half cycles of the mains. The thyristor cathodes are connected to logic 0V, which means that this is **not** at the same potential as earth and no connection should be made between this and the case, which is earthed for safety.

CONSTRUCTION

As a wrong connection could have dramatic results, it is suggested that construction is done in stages, and each stage checked. The best order is logic power supply, multivibrator, counting and decoding i.c.s, mains rectifier and thyristors, and zero crossing pulse generator.

The Veroboard cuts and component layout are shown in Fig. 3 and the accompanying photographs. Individual constructors will no doubt modify this as they see fit to accommodate components of differing

dimensions.

system showing the four columns of bulbs in dashed line. Note that the live side of the mains supply is connected to the logic OV line through D4 and thus care is required in ensuring insulation of the circuitry from the case and controls

. Check the board carefully for shorts between adjacent strips and for strips not cut in the right place. Note that there are not always seven cuts underneath i.c.s as sometimes it is necessary to join the opposite pins.

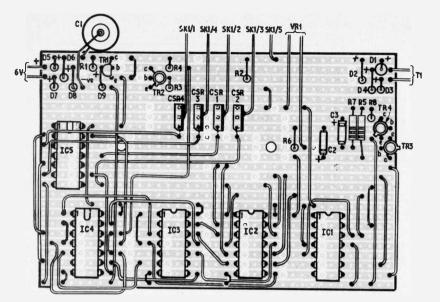
The thyristors used in the prototype had their anodes connected to their heat sinks. The wires to the output socket were soldered to the heat sinks rather than to the anode wire on the Veroboard, as this was more convenient.

The five-way connector used for the output should be as non-standard as possible for safety. A five-pin DIN would work but is almost bound to be connected to audio equipment at some time.

THE PIPE

The bulbs used in the prototype pipe were 8V, 150mA, 11mm diameter, Vitality type no. G537. A supplier for these is mentioned in the components list. They are a tight squeeze in the pipe, but once installed, the pipe may be bent round fairly sharp corners without danger of them breaking.

If a smaller bore pipe is being used, there is an electrically similar type, the 676W/E, which is wire ended and much smaller. They are, however,







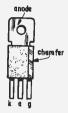


Fig. 3. Component layout and Veroboard cuts and interconnections for the Light Pipe showing connections to unmounted components

more expensive, and come in packs of 200. These are available from the same supplier.

As the lamps are somewhat inaccessible in the pipe it is important that there are as few failures as possible. The lamps are considerably underrun to increase their life. This also reduces light output and hence temperature. The pipe should be mildly warm after a few minutes running, but there is no danger of the plastic melting.

CONSTRUCTION OF THE PIPE

First, solder the bulbs into four chains of 36 bulbs. There should be 8in between each bulb for an overall length of eight yards. The connecting wire should be as thin as possible and multistranded. Some sort of jig to hold the bulbs while soldering is of great help in speeding up the operation.

The chains should be tested on the mains at this stage to find any blown or broken bulbs, or dry

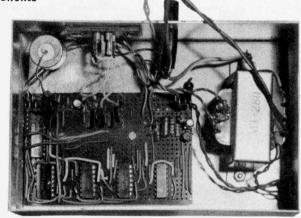
joints.

repeated 36 times along tube

Tape leads together at intervals

Chains to PL1/1, V2,1/3 and 1/4

Fig. 4. Sketch showing the method of tying the light bulbs to draw them through the pipe. Note that particularly with the wire-ended bulbs the taping of the various conductors between each bulb is important in protecting the bulbs during installation



The four chains should be laid out so that there is 2in between each bulb, and taped together at intervals with a fifth common wire. The pipe is prepared for insertion by blowing talcum powder down it.

This reduces friction which otherwise might impose too great a strain on the bulbs during insertion.

To thread the lights, a piece of string must first be put through the pipe, using a piece of iron or steel and a magnet to pull it through. The string is used to pull the light chain back through. The bulbs should be pulled through with great care in order to avoid damaging them where the lead-out wires emerge from the envelope if the 676 type is used.

It is easier to fix the string at the far end of the pipe, and pull the pipe over the lamps, with liberal addition of talcum powder if it is difficult to pull.

The common wire should be joined at one end to the four other wires and held at the end by tying a piece of wire round the joint and tying this to a piece of plastic which is glued over the end of the pipe.

The wire leading to the pipe is two yards of fivecore (four-core + screen audio will do), with a suitable plug, of which any metal parts should be well

insulated from the pins. The joint between it and the pipe is hidden by wide insulating tape wrapped round the pipe. The far end is similarly covered to improve appearance.

FAULT FINDING

In the event of a lamp failure, it is easier to cut into the tube and replace the lamp in situ rather than pull out all the lamps and deal with it externally.

The bulbs which have gone out should be marked on the outside of the tube with a felt pen. Then by pushing a pin through the tube it is possible to make contact with one side of the bulb and the continuity between it and the common line can be tested. If the first test is done in the centre, further tests can be made to subdivide the faulty section until the blown bulb is found.

Whilst cutting the tube might seem drastic it is certainly simple and providing a sharp knife or scalpel is used to give a clean cut it is not difficult to re-glue the cut faces using one of several fast

COMPONENTS

Resistors 100Ω R1

R2 $470k\Omega$

R3 4.7kΩ

R4 $1k\Omega$

R5 1.2kΩ

R₆ 3-3kΩ

R7 $10k\Omega$ R8 1.2kΩ

Potentiometer

VR1 220kΩ linear with double pole mains switch

Capacitors

640μF 10V electrolytic C2, C3 4µF 10V electrolytic (2 off)

Transistors

TR1-4 BC108 (4 off)

Diodes

D1-4 1N4005 (4 off) D5-8 1N4001 (4 off)

D9 5.6V Zener

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SN7473 IC2-5 SN7400 (4 off)

Thyristors

CSR1 to 4 CRS1/40 (1A, 400V)

Miscellaneous

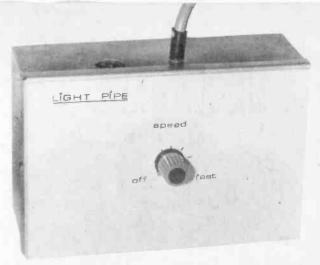
144 type G537 8V 0-15A bulbs (Townsend-Coates (Vitality) Ltd., Coleman Road, Leicester, LE5 4LP. Valiant Electrical Wholesale, 20 Lettuce St., Fulham, London. Farnell Electronic Components)

8 yards green polythene tube (o.d. 2.3cm, i.d. 1.6cm) (Transatlantic Plastics, Surbiton) Aluminium box type AB13

Veroboard 24 strips × 36 holes Transformer Eagle MT280 6V,

Suitable 5-pin socket and plug assembly. 1A fuse and fuse-

holder



setting adhesives now on the market such as Thixofix or Araldite. Also the cut is usefully 'L'-shaped rather than through the whole pipe.

DEVELOPMENTS

The direction of apparent motion may be reversed by reversing the anode wires to CSR1 and CSR3, and if desired a switch may be included for this

There is no reason why several pipes may not be run off the same controller as the thyristors are rated at 1A and each chain takes only 150mA. It could be arranged to flash coloured lamps as well as the light pipe, provided the current is kept well within the ratings of the thyristors. 3A types could be substituted with no circuit modifications if a high power output is needed.

Any increase in demand for current will of course affect the choice of diodes D1 to 4 and these will need uprating if demand exceeds 1A.

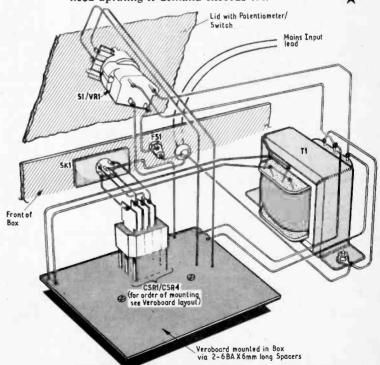


Fig. 5. Interwiring diagram



BYJ.S. HAGGIS

NE requirement for greenhouse owners whether amateur or professional is a system for keeping leaf surfaces moist. This is achieved by spraying the leaves, at intervals, with a fine water mist. The timing of the intervals depends on the rate of evaporation at any one time. This article describes the principle, theory, and construction of a device that will automatically control this process.

The system uses a sensor which consists of two electrodes placed approximately 1 in apart and cast in and Araldite block. The surface of this block is machined flat so that when it is sprayed with water the two electrodes eventually become bridged with a blob of water. If no further spray is directed at the surface, evaporation soon starts to reduce the amount of water bridging the electrodes, until finally the bridge is broken. The time taken for this to happen depends on the rate of evaporation. On a hot summer day the time will be short, whereas on a damp cloudy day the time will be longer.

PRACTICAL CIRCUIT

The making and breaking of the sensor circuit can be used to operate the electronic circuit of Fig. 1. At first sight this might seem rather complex for such a simple operation, but as will be seen later the system allows for quite a degree of control, coupled with sound operation and reliability.

The object of the circuit is to actuate a standard mains operated solenoid water valve to feed water to a mist propagator when required, and to do this with no moving parts except the solenoid valve, i.e. relays, thus eliminating any trouble from corroded contacts which can easily happen in the greenhouse environment.

Electrode making and breaking is sensed by TR1 and here one of the prime requirements is to keep the current that flows through the sensor, when covered with water, as low as possible to reduce the effect of electrolysis which would soon fur-up the two electrodes.

With no water bridging the electrodes TR1 base is open-circuit therefore the collector is almost at supply potential. As soon as the electrodes are bridged the base goes high turning TR1 on, the

collector then going to 0 volts.

At this stage it could be argued that no further circuitry is required other than a relay or thyristor to switch the mains. This would be fine if water behaved to our liking, but in fact this is not so. Once the sensor has been exposed to the atmosphere for some time the surface becomes slightly greasy. This causes the water droplets to have greater surface tension which means in practice that the sensor must be fully covered with water before the spray is stopped or intermittent operation of the circuit will take place, settling down once the surface has accumulated enough water.

TIMER

However, to overcome this problem and to give more choice of control a timer has been added. This takes the form of the now well-known 555 i.c. (IC2) discussed in depth in June 1973 PRACTICAL ELECTRONICS.

The 555 is used in the monostable mode and thus the trigger pulse must be shorter than the timing period. Differentiation of the input signal is achieved by R4 and C3. In order to provide a sharp pulse edge a Schmitt trigger is required and this also provides iitter-free operation.

The Schmitt used is a monolithic integrated, 14 pin package, type SN 7413 (IC1). The package contains two identical triggers. Each circuit functions as a four input NAND gate. All four inputs are held high by internal resistors and in this condition the output is held at logic 0. Any one input going to logic 0 sends the output to logic 1 level, i.e. 5V.

If any constructor wishes to omit the timer circuit and chance any intermittent switch-off, he can do so by leaving out the timer package and associated

components.



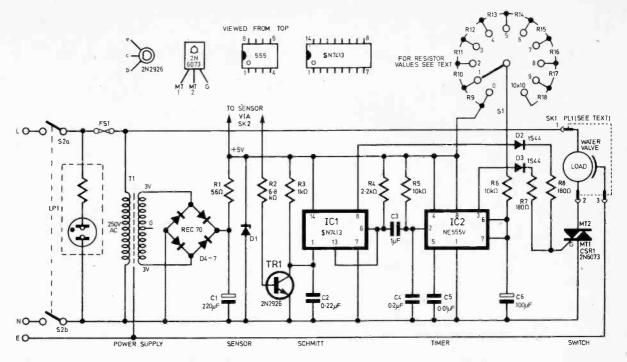


Fig. 1. Circuit diagram of the complete plant moisture control system. Note that the earth line is connected to the case of the control valve if possible and to any metal case parts if these are used

CIRCUIT DESCRIPTION

As can be seen, the power supply is not special. The 5 volt rail is Zener stabilised whilst the mains transformer T1 is an R.S. Components Ltd., sub. min. 3-0-3V device rated at 200mA. The total circuit consumption is 45mA. Even so the transformer gets quite warm but long tests have been given to the prototype with no ill effect.

As stated earlier, when the sensor X1 is bridged with a blob of water the collector of TR1 is at OV. This state is fed to that trigger input whose output under these conditions is high. The requirement is to have a high output on C3, because the

timer is started by a falling voltage.

When the blob of water has evaporated sufficiently to break the contact between the electrodes of the sensor, the base of TR1 becomes open circuit and the collector goes high. Thus the output of the Schmitt goes low and starts the timer, the output of which goes high, to approximately 5V. This is applied through a resistor to the gate of CSR1 thus turning it on.

The time the output of the timer remains in this state is determined by the setting of S1 which selects

the series resistors R9 to R18.

The electrodes can be bridged again with the spraying water with no further effect on the timer. Once the time sequence is ended the circuit is set for the next sequence when evaporation dictates.

A brief look at the circuit diagram will reveal a direct connection through diode D2 and resistor R8 from the output of the second Schmitt to the gate of CSR1.

This has two functions, the first being if the operator wishes the sensor to control the "off" time as well as the "on", and accept the possibility of an unstable switch-off, he can do so by turning the time

COMPONENTS . . .

Resisto	rs		
R1	56 Ω	R6	10k Ω
R2	6.8k Ω	R7	180 Ω See text
R3	1kΩ	R8	180 Ω See text
R4	2.2k Ω	R9 to R18	See text
R5	10k Ω		
All ½ v	vatt carbon		

Capacitors

220 µF elect. 40V printed circuit type

0.22µF disc C2

1μF non-elect 63V Wima C3

C4 0.2μF disc

C5 0.01 µF disc

100μF electrolytic 63V Wima

Semiconductors

2N 2926 TR1 CSR₁ 2N 6073 IC1 SN 7413 IC2 NE 555V

4.7V, 1.3W Zener BZX61 series or D1

IS2000A, IS7000A series (Doram)

D2 and D3 1S44 (2 off)

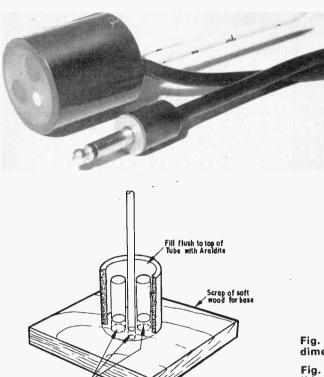
Rec 70 (800mA 400V) Doram (4 off) D4 to D7

Miscellaneous

T1 3V type 200mA. Doram SK2/PL2 Jack and socket min.

Case Sarel Ref. No. 308 from Hawnt and Co. Ltd.,

Pritchett Street, Birmingham B6 4EN. Mains cable 3 core P.V.C. Circuit board. Rubber flex connector (Woolworths). Neon indicator 250V miniature. Knob and dial skirt 'C' ∄in. Doram. Midget rotary switch 1 pole 12 way. Miniature toggle ON/OFF switch. Mains fuse and socket. Materials for sensor. Solenoid valvesee text.



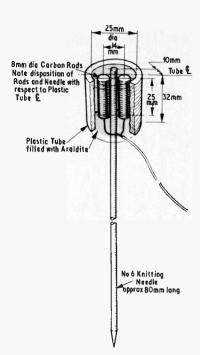


Fig. 2. A completed sensor (above) showing the dimensions in mm and simplicity of construction

Fig. 3. Manufacture of the sensor is also simple using this jig made up from a bit of scrap wood (left)

control to its minimum. Thus when the water breaks the contact the collector, as before, goes high, and starts a timing cycle which with the control turned to minimum, is only in the order of 1s and therefore of no consequence.

Make holes in base for placement of Rods and Needle

But the second Schmitt output volts are applied direct to CSR1 gate and this turns on CSR1 for as long as the electrodes remain open. Each Schmitt within the package inverts therefore the output from the first Schmitt must be fed into the input of the second which is in turn used for direct control of CSR1.

Without the D2, R8 connection and the timer interval inadvertently set too short to allow full coverage of the sensor with water, the timer would be triggered as before thereby turning on the spray. But if the spray were to cease before the sensor was properly bridged, no further action could be taken by the circuit. In fact if man did not intervene the circuit would remain stuck in this state for good. But with the direct connection from the Schmitt this cannot happen. The two diodes are to prevent any possible interaction.

SENSOR

The sensor X1 consists of two ½ in carbon rods cast in Araldite at ½ in centres (see Fig. 2).

The design of the sensor is quite critical. The relationship between the total diameter of the sensor, the diameter of the carbon rods, and their separation distance is important. Because of surface tension the blob of water that forms on the top of the sensor can roll about the surface and avoid the

carbons and even roll off the surface if the design is not correct.

SENSOR CONSTRUCTION

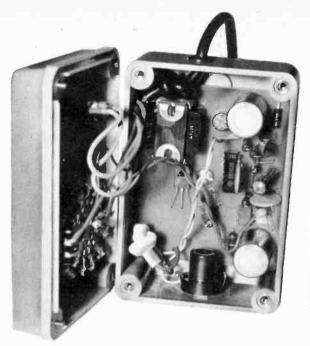
The sensor is best cast in a piece of plastic pipe with a lin inside diameter. This has two advantages, the first being ease of casting and the second and more important being that the plastic edge surrounding the sensor surface tends to prevent the blob of water from rolling off.

The two carbon rods can be obtained from any used-up U2 type battery, the metal cap ends being used to solder the wire connections.

Cut the two carbon rods at $1\frac{1}{4}$ in long from the metal cap end. Drill two holes $\frac{5}{16}$ in dia. at $\frac{5}{8}$ in centres, $\frac{1}{4}$ in into a piece of scrap softwood. Drill a third hole to one side, as indicated in Fig. 3 to accommodate a No. 6 plastic knitting needle.

Push the two carbon rods into the holes as far as possible. Acquire a 4½ in length of No. 6 knitting needle (sharp end if possible) and insert the blunt end into the third hole as far as it will go. Next place a 1½ in length of 1 in inside dia. plastic tube over the carbon rods and needle. All is now ready to attach the wires. Any twin cable with a plastic outer sheath is suitable. Solder one wire each onto the metal caps of the carbon rods. Ensure that the cable sheath is below the edge of the tube so that it will be well covered with Araldite.

When all is ready mix the Araldite resin as directed and heat until it runs quite freely, then pour it into the plastic tube ensuring that air bubbles are not trapped. Rotate the tube a little and position correctly round the carbons. Leave to set over night. One packet of Araldite is sufficient.



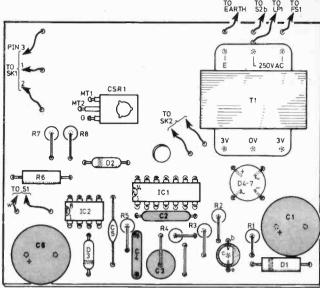


Fig. 4. Component layout on the prototype printed circuit board

Once the Araldite has hardened the timber can be split from around the ends of the carbon rods. Saw the carbons and needle flush across the surface of the sensor and then sandpaper the surface flat, finally finishing with a very fine sandpaper. Wipe the surface with methylated spirits. The sensor is now ready for use.

CIRCUIT CONSTRUCTION

In the prototype all components were mounted on one printed circuit board, including the mains transformer as shown in Fig. 4 and using the circuit layout of Fig. 5. If any constructor wishes to alter this arrangement there is no problem, the layout is not critical other than the necessity to adequately separate adjacent conductors carrying 230V a.c.

It might be noted here that the whole circuit is connected to the mains neutral line and care should be taken when handling the circuit board when switched on as full mains potential is on the board. Preferably the board should not be handled at all when power is connected.

The board is housed in a Sarel box measuring $4\frac{1}{4} \times 2\frac{1}{4} \times 2$ in. The Sarel boxes are made from a tough plastic and are waterproof which makes them safe when containing mains in possibly damp conditions.

The three core mains lead enters the box through a tight fitting grommet and may either be wired direct into a fused distribution box or connected to the mains via a fused standard square pin 13A plug. It is suggested that this cable be black.

The cable to the load, in this case a water valve, again leaves the box through a tight fitting grommet and is terminated with a three pin rubber flex connection to which the water valve may be connected. It is suggested that this cable be white to eliminate any possible confusion.

The sensor X1 is plugged into the box using a normal jack plug PL1 and socket SK1. To enable the circuit to be tested or as a convenient way of

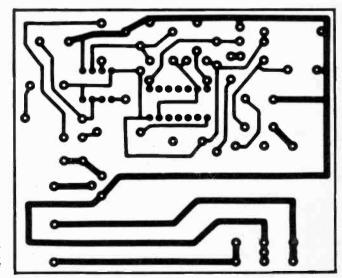
operating the water valve manually, the jack socket is wired such that when the jack is withdrawn the base of TR1 is left open circuit, thereby turning on CSR1 for as long as the jack remains out of the socket.

C2 should be mounted as closely as possible to the Schmitt input. This is best achieved by soldering it directly onto the package pins.

SETTING UP

First ensure that the circuit has the correct d.c. voltage, 5V, and if not change the value of R1 accordingly. For most practical purposes a maximum time value of 100 seconds in 10 second steps

Fig. 5. Printed circuit master as used in the prototype Leaf



is quite adequate. But if longer or shorter time intervals are required it is only necessary to change the value of C6, and/or the values of R9 to R18 on the rotary switch S1.

S1 is a midget rotary 12 way switch set to give 11 positions and assembled with knob and dial skirt "C" in order to give indications from 0 to 10.

With C6 at $100\mu\text{F}$ and a maximum time of 100 seconds required, the total value of R9 to R18 is approximately $500\text{k}\Omega$. This will vary according to tolerance spread of different condensers.

In order to test and set-up the circuit, connect a 60W mains bulb in the load position SK1, pins 1 and 2. This gives clear indication of the operation of the circuit. Then stand the sensor up and cover the sensing surface with a small piece of very wet blotting paper. Turn the rotary switch to the 0 position, plug in the sensor, apply the mains and switch on.

If all is working correctly the bulb should light up as soon as the wet blotting paper is removed and remain lit as long as the paper is off the sensor. Replace the blotting paper and the light should extinguish about I second later.

If when the bulb is lit. it appears to flicker and not to be at full brilliance it is because CSR1 is not switched fully on. This could happen because of variations in sensitivity of individual devices and can be rectified by reducing the value of R7 and R8.

TIMING RESISTORS

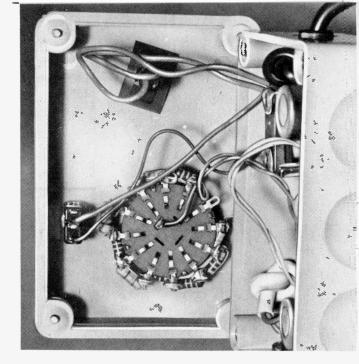
If all is well up to this stage it now only remains to fit the individual timing resistors around \$1. Solder into the first position a 47k\O resistor and select I on the dial (remember all these numbers on the dial are ×10). Then with a suitable time piece handy, remove and replace the blotting paper and note the length of time the bulb stays illuminated. With preferred resistor values it may be difficult to time precisely, but times within 10 per cent can be achieved. If for instance, the 47k\(\Omega\) gives 11 seconds instead of the required 10 seconds, then in the next position solder a $33k\Omega$; this should then bring the 20 second timing to about 18 or 19 seconds. It is possible by carrying on in this manner to stay quite close to the required timing. More accuracy can be gained, if required, by making use of resistor tolerance spread.

The repeatable timings of the circuit are very good and this is what matters most. It does not matter much that a time of 62 seconds is given when 60 is wanted as long as it repeats.

INSTALLATION

Securely mount the control box and connect the circuit to the mains through a fuse box or a 13A square pin fused plug fitted with a 3A (or thereabouts) fuse. Connect up a suitable mains operated water valve for the spray system.

CSR1 is rated at 4A and this load should not be exceeded or damage will occur. There is little advantage in fitting a fuse because in general CSR1 will blow before a fuse. If it is required to switch greater loads it will be necessary to select a different CSR1.



Stick the sensor spike into the soil at a convenient spot amongst the plants to be sprayed, but not covered by them. Wipe the top of the sensor with some methylated spirits to remove any grease. Select position 0 and switch on noting the length of time required for the sensor surface to become covered and the system to switch off.

It is unlikely that subsequent operations will take as long as the first because the subsequent operations will not be starting from a dry surface, but an idea of the time required can be gained. The unit can now either be left on "0" and control its own switch off time as stated earlier, or set to a time interval as suggested by the initial timing. Experience will soon enable a useful setting to be found.

MAINTENANCE

The only maintenance required is an occasional wipe of the sensor with meths.

If at any time it is required to test the unit it is only necessary to remove the sensor jack and replace it again and the system will operate for as long as the selected time interval set.

SOLENOID VALVE

The choice of valve used depends on the water pipe system in use, water pressure and availability. The author used one from C. W. Wheelhouse & Sons, 9-13 Bell Road, Hounslow, Middx. which had in B.S.P. fittings.

The EVJD10 and the EVD15 are 1 in B.S.P. male and female fitting valves from Danfoss (London) Ltd., 6 Wadsworth Road, Perivale, Greenford, Middx. Both cover most water pressures which will be met and are available from stockists at around £12.

A d.c. operated valve could be accommodated by using a transformer to supply the correct voltage and a suitable rectifier.

MERCURY AGAIN

The second visit to Mercury by Mariner 10 confirmed the previous assessment generally. The second fly-by was 50,000km. More than 500 high resolution pictures have been received, all of excellent quality. This pass was entirely on the sunlit side of the planet and covered some areas that had been checked on the first pass in March last year. There were no startling new discoveries.

The hemisphere observed showed that the planet has only one large impact basin. This area known as the Colaris Basin was on the terminator at the first pass. The floor of the crater would seem to be about the size of the Mare Imbrium

of the Moon.

Many craters seem to have a central peak which suggests that the terrain was extremely soft at the time of impacts. There are also many craters within craters. A number of rayed craters exist and there is one large area near the south pole with the rays extending out through other craters. The south pole is inside a large crater.

There seem to be two distinct types of configuration of craters, those which are concentric and those which are in long chains. The concentric craters seem to be of the order of 200km in diameter.

Another feature is the large number of random scarps hundreds of kilometres long and of heights of the order of one to three kilometres. The stresses shown are compressional and the result of the compression produces these scarps. They appear to be random in direction and do not follow any particular pattern. The scarps appear to be antipodal to the Colaris Basin and a similar situation is indicated on the Moon in relation to Mare

It would appear that the condition in which Mercury is now found fits in very well with the current theories of early development of the Solar system. This means that there was a huge infall of large bodies on the emerging planets followed by a period of high internal heating with volcanic activity. This takes care of the changing surface features leaving the primordal conditions hardly discernible.

OPTICAL NAVIGATION IN SPACE

Another important experiment carried out by Mariner 10 was to test the optical techniques for navigation. These techniques are vital to the missions to outer planets. The need to be able to dodge the satellites of Jupiter, Saturn and



Uranus is imperative because of the sparse data as to the ephemerides of these bodies.

Previous navigation of deep space missions have relied on Earth-based radio measurements. Mariner 10 carried out a real-time experiment as it approached Mercury. Over a hundred pictures were taken to show the angular between Mercury and the nearby (optically) stars. The experiment was successful and this means that spacecraft on long distant missions will be able to continuously monitor the space ahead and therefore automatically adjust course.

Provided Mariner 10 can limit its fuel for modifying its course and not suffer another distraction because fast particles interfere with the sensor which is locked on Canopus, it should be possible for the spacecraft to make another pass of Mercury in 1975.

MORE FROM SKYLAB

Adding to the preliminary reports that have appeared about the effect of weightlessness and the general ability of the astronauts to be in a condition to carry out their tasks, some particular findings are now available. These were released at symposium arranged by the American Astronautical Society.

Altogether there were 171 days of free orbital flight during the Skylab mission and not only has it been demonstrated that man can adapt indefinitely to the weightless environment, it has also shown that provided there has been adequate exercising during the mission the return to normal gravity presents no problem. The state of the individual readily adapts to the return to Earth with no resultant effects. This settles the problems that were thought to exist for long missions into space.

Many ways of overcoming the weightless condition have been investigated and these are no longer required. All that is needed for an astronaut is a programme of physical exercises for 90 minutes each day during the flight.

ASTRONAUT PERFORMANCE

In each of the Skylab missions there was a progressive improvement in astronaut performance. This was due mainly to the increased exercise taken. The first crew were so occupied with the various technical problems that arose; little time was available for an intense study of the exercise aspect. Even when the extra vehicular activities continued to six hours at a stretch there were no "clinical" events to record.

No signs of heart deterioration appeared but there was the usual loss of red cell mass, this reduced as the exercising extended and the flights were longer. Thus the first mission crew lost 14 per cent, the second crew 12.3 per cent but the third crew loss was down to 6.8 per cent. As on the Apollo missions there was a wide variation between individuals.

Post recovery was better in the case of the last mission of 84 days. The astronauts on the last mission lost least weight. Most of them gained in height, this averaged about an inch. There was a shift-ing of the body fluids. Some of the crew experienced slight malaise for the first two or three days. However, all got their "space legs" in a day or so and thereafter were immune.

The calorific requirements proved to be the same or thereabouts as their normal routine on

Testing the ability of these men as observers of the Earth from space, an analysis made of the observations and the ability to carry out the tasks allotted showed that in 850 observations and some 2.000 hand held camera shots, during the 84 days' mission covering 83 widely varied categories, led to these conclusions. The ability to recognise objects and patterns, to integrate these observations from a wide range of aspects and lighting angles, to reason, to make selected observations and describe them brings a new dimension into play. This ability transcended anything that could be programmed or made automatic.

NOBEL PRIZE FOR CAMBRIDGE RADIO ASTRONOMERS By Frank Hyde

THE announcement that the Nobel Prize for Physics had been awarded to Professor Sir Martin Ryle FRS, now the Astronomer Royal, and Professor Anthony Hewish FRS, is a fitting reward for the work of these two quiet and unassuming men from the Mullard Radio Astronomy Observatory. It is a far cry from the angle iron and wire structures, which began a series of programmes more than two decades ago, to the 5km radio telescope now in operation at Lord's Bridge near Cambridge.

The official citation for the award ends with these words "for their pioneering research in radioastrophysics: Ryle for his observations and inventions, in particular the aperture-synthesis technique and Hewish for his decisive role in the discovery of pulsars"

Ryle, together with Hewish and Graham-Smith, were the original team. Hewish has stayed on but Graham-Smith moved to a Professorial Chair at Manchester and on the 1st January 1975 takes up the position of Director of the Royal Greenwich Observatory.

INTERFEROMETRY

The techniques of interferometry formed the basis of the work at Cambridge. The first telescopes were simple interferometers but very soon the phase switched inter-ferometer came into being; the details of this were published about 1950 by Ryle and Vonberg.

From the early beginning the resolution obtained by these methods was better than the single aerial systems. The interferometer developed into a number of variations and gave rise to the more advanced technique of aperture synthesis for which Cambridge is famous. Digitation of observations led to a number of advances.

The first aperture synthesis aerials consisted of one long corner reflector and one movable one about forty feet long. In those days it was quite a sight to see the small reflector being carried to its new position by the observers and technician. This was followed by a more sophisticated cylindrical parabola driven in attitude by synchronous motors and the smaller complementary aerial fitted to a bogey on a railway system. Here the long aerial was set up east and west and the smaller one could travel on the rails north and south.

APERTURE SYNTHESIS

The technique of aperture synthesis can be described simply as two apertures moving relative to each other in such a way that they sweep out a narrow ring of large diameter. The apertures are the aerials and the rotation of the earth varies their orientation such that seen from the sky, one aerial appears to trace out an ellipse relative to the other. The interference patterns which are made by superimposing the signals of the two aerials are then synthesised. From this a chart of the structure of a source can be deduced.

The next major step was the One Mile Telescope which came into operation at the end of 1964. This was the first complete aperture synthesis purpose-built telescope. The sensitivity was extremely high. It was able to detect faint objects near the edge of the observ-

able universe.

At that time the cosmologica! debate was in full flow. The first results from the new telescope showed quite clearly that the count of the sources at great distances (and therefore very old) was less than the number required by the "steady state" theory, but consistently supported the "big bang" theory. For most of the theorists this fact, together with discovery of the microwave background, which appeared universal, marked the end of support for the "steady state" protagonists.

MAPPING RADIO GALAXIES AND QUASARS

Leading on from the far-distant-source discoveries. the next successful objective was the mapping of the radio galaxies and quasars. The properties of the galaxies were of particular interest because they are enormous wasters of energy and among the largest objects known.

The detailed mapping of the galaxies and quasars enabled some conclusions to be drawn as to their birth, evolution and final demise. In this area Ryle not only made the initial discoveries but also showed that the exploding galaxies threw out on each side large clouds. These clouds appear to interact with the inter-galactic medium. These great radio clouds seem to continue to receive energy from the optically observable nucleus of the original galaxy or quasar for millions of years.

In 1964 at the International Astronomical Union meeting in Hamburg, Anthony Hewish released the details of his method of finding quasars by their scintillation. A quasar is to the radio telescope a small object of high intensity. The interplanetary medium gives changes of density with the same effects as the atmosphere on light coming toward the Earth. The small apparent diameter of a quasar is comparable with the variations. In consequence a quasar is revealed by the amount of scintillation that takes place.

The preliminary work on this problem was undertaken with three stations, Cambridge, Thetford, and Clacton, roughly in an equilateral triangle. The particular quasar studied was 3C 48. This showed nearly 50 per cent scintillation. The pilot experiment was successful and the triangle was extended between Cambridge, Jodrell Bank,

and Malvern.

DISCOVERY OF PULSARS

It was during the testing of the 18,000 square metre aerial at Cambridge that Jocelyn Bell first noticed the regular pulses. Later observations showed that the regularity and accuracy of the pulses was more reliable than any clock available. The team went to great lengths to establish what these pulses really were and what mechanism was involved. It became clear that what had been discovered were the playthings of the theoretical astronomers, the neutron stars. This discovery was extremely exciting, opening up as it did new possibilities in gravitational physics, the behaviour of very dense matter, and the effect on radiation physics.

SUPER-SYNTHESIS

The reward for enterprise had already been given by the Science Research Council in approving the setting up of the 5km aerial with its eight parabolas on the site of the old railway axed by Beeching. This system of aerials went one step further than aperture synthesis. It reached the stage of super-synthesis.

The principal task was to be the examination at the new high level of resolution, of quasars and radio galaxies. This was an opportunity to resolve some of the problems. In some of the observations there were more than two radio clouds associated with the explosions that took place. There were also bridge-like links between the areas of activity. The radio emission may come from the highly charged particles moving at speeds near that of light and trapped in a tangled magnetic field. They may well be accelerated by the gravitational collapse of groups of stars near the centres of visible galaxies.

All these problems and many others should respond

to the high resolution of the 5km telescope. Since 1972 the telescope has already shown that its inception was

more than justified.

THE MAKING OF HISTORY

The history of radio astronomy as made at the Mullard Radio Astronomy Observatory at Cambridge includes many activities which need a whole volume for description, but the story will go down in history of the time when two men, mainly responsible for that history, were honoured while still making it.



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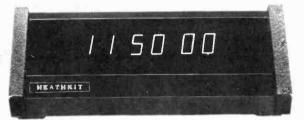
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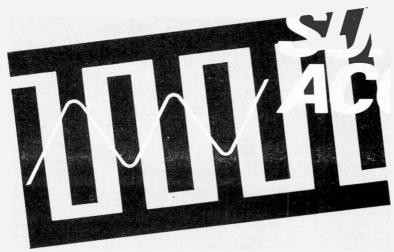
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HEATH



So you've never heard of surface acoustic waves. Well you soon will. Progress has been astonishing in the last five years. Devices have moved from starry eyed physicists' dreams to commercial evaluation in electronic systems—a real case of fantasy to fact.

Imagine you are standing on your favourite holiday beach looking out to sea; a yacht rocks peacefully at the end of the breakwater whilst the waves are gently lapping the sand at your feet. You are witnessing the analogy of surface acoustic waves (S.A.W.).

Transfer the sea waves to the surface of a piece of flat polished crystal and you have S.A.W. Thin metal films deposited on the crystal surface enable the transfer between electrical and S.A.W. energy to be effected. By manipulating these waves on the crystal surface, oscillators, amplifiers, signal processors and delay lines have been constructed.

HISTORICAL VIEWPOINT

Lord Rayleigh is accredited with the identification of the surface acoustic wave. In 1885 he described waves travelling along the earth's surface after an earthquake, and subsequently a great deal of information has been gathered by seismologists.

It was not until the early sixties that the propagation of high frequency sound waves through a solid crystal was demonstrated. The development of a means of conversion between electrical energy and S.A.W. energy, the Interdigital Transducer (I.D.T.) in the mid and late sixties, meant that the breakthrough had come. S.A.W. exploitation was here.

THE NATURE OF S.A.W.

Surface Acoustic Waves are only one in a family of wave motions identified in crystalline materials. The property of all these waves is that of transferring acoustic energy from one part of the crystal structure to another.

In crystalline materials the particles are arranged in an orderly lattice type of structure, each particle being held in place by an elastic force generated between itself and its neighbours (imagine a lattice of billiard balls coupled by pieces of rubber band).

The longitudinal wave travels through an elastic material by alternately expanding and compressing the crystal lattice. This is definitely a bulk wave not found on the surface of the crystal. The second type, the shear wave, vibrates the lattice at right angles to the direction in which the wave is travelling. The layers of the lattice slide up and down past each other. Again a bulk wave not found on the crystal surface.

The third type is the S.A.W. (Fig. 1a). It is a combination of the longitudinal and shear waves and is only associated with the surface of a crystal. To tie up the analogy with the waves on the surface of the sea, mentioned earlier, it is interesting to note that the motion of a particle travelling in a S.A.W. is also retrograde elliptical (Fig. 1b).

To digress for a moment; bulk waves occur in the familiar, so called, crystal oscillators which are widely used for highly stable frequency sources in, amongst other things, communications equipment.

It is easy to appreciate early fears that the frictional forces generated by this mechanical vibration of the crystal lattice would absorb the acoustic wave energy excessively when the frequency of oscillation got too high. Fortunately crystals of quartz and lithium niobate have been found to transmit frequencies from 30MHz to 10GHz acceptably.

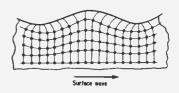




Fig. 1(a). Surface Acoustic Waves distort the crystal lattice as the wave moves along. A particle near the surface exhibits retrograde elliptical motion (b) directly analogous to waves in water

THE SPEED OF S.A.W.

We now come to one of the outstanding properties of S.A.W.—the velocity with which the waves travel along the crystal is around 10⁵ times slower than the velocity of electromagnetic waves, i.e. the velocity of light. In other words an electronic signal that would occupy a cable one mile long could be contained on the surface of a piece of crystal only half an inch long. This means that the time taken for the signal to travel one mile in the cable would be the same as the time taken for it to travel just half an inch on the crystal.

It is now possible to glimpse how a delay line of incredible compactness might be constructed. But first a means of transferring the electrical signal to the surface of the crystal (into a S.A.W.) is needed. Ideally the process should be reversible so that the electrical signal can be recovered after it has been delayed. A transducer is required. Fig. 2 shows a delay line with input and output transducers.

THE TRANSDUCER

The transducing action for converting electrical energy to S.A.W. energy occurs in two parts. The first stage is a conversion of the incoming electrical signal to an electrical field which varies in strength and polarity as the incoming signal.

The second stage takes advantage of the property of piezo-electric materials to mechanically vibrate

in sympathy with an applied electric field.

As mentioned above, suitable crystals for the transmission of S.A.W. are quartz and lithium niobate, both of which are also piezo-electric. Thus, all that is necessary to excite a S.A.W. is to apply a suitable temporally varying electric field to the piezo-electric crystal substrate which will also transmit the S.A.W. This is done using an Interdigital Transducer (I.D.T.).

THE INTERDIGITAL TRANSDUCER

Imagine the fingers of your left hand interleaved with your right and you have the form of an I.D.T. This shape is laid down on the crystal substrate as a thin aluminium film.

The incoming electrical signal is applied with each hand (in analogy only of course!) acting as the terminals. The electric field will be generated between the finger of each hand since there will be a time dependent potential between them due to the

electric signal (Fig. 3).

The spacing between each of the fingers is equal to half the wavelength of the S.A.W. and the width of the fingers is typically a quarter wavelength. The wavelength is determined from a precise knowledge of the velocity of the S.A.W. on that particular crystal substrate and the frequency of the incoming electrical signal.

The aperture of the transducer (see Fig. 3) determines the impedance seen by the incoming electrical signal and where possible is made to match the line impedance, e.g. 50 ohms. This typically means an

aperture of 20 to 100 wavelengths.

The transducer described generates a bi-directional S.A.W. The bandwidth of this I.D.T. is inversely proportional to the number of finger pairs in the transducer (see Fig. 3). The inverse of this same process of S.A.W. generation is used to detect S.A.W., i.e. to generate an electrical signal from an incoming S.A.W.

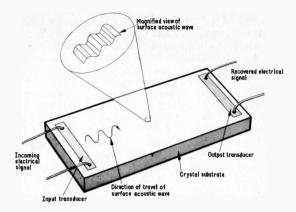


Fig. 2. A delay line consists of a crystal substrate with transducers separated by a distance depending on the delay required

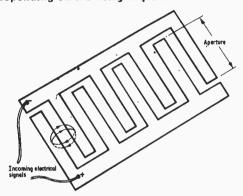


Fig. 3. An interdigital transducer is used to induce the S.A.W. into the crystal. At the instance depicted the polarity of the signal and the lines of force between two points of a finger pair are as shown. An alternating electric signal produces an alternating field.

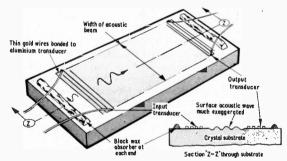


Fig. 4(a). Detailed structure of a delay line showing I.D.T.s and wax which absorbs waves so preventing unwanted reflections from the ends of the crystal. (b) shows the delay line in cross section.

DELAY LINES

Delay lines are perhaps the most fundamental S.A.W. devices. Their function is simply to provide a delay between the receipt of a signal and its onward transmission, leaving its form unchanged.

They comprise of an I.D.T. at either end of a crystal substrate. The length of the crystal determines the delay (see Fig. 4).

Since the I.D.T. is bi-directional, half the S.A.W. energy is radiated in the wrong direction and if not stopped would reflect back from the edge of the crystal thus causing an interfering signal. Thinking of the sea shore analogy, mentioned above, imagine the waves hitting the wall of a cliff or promenade and being reflected back out to sea.

Black wax is used to absorb these unwanted S.A.W. It is painted on the crystal surface and the effect can be likened to the sand of the sea shore which tends to dissipate the waves' energy.

Omni-directional transducers can be constructed but these have only two thirds the bandwidth of the bi-directional transducer. The delay possible with lithium niobate, for instance, is 2.88 microseconds per centimetre.

FABRICATION

The method of construction is common to most forms of S.A.W. device. Having selected the crystal required, cut and polished it, the I.D.T. thin metal films are deposited using conventional integrated circuit techniques.

Electrical connection to the I.D.T. is made via extremely fine gold wires bonded to the metal film. These wires are typically one to two thousandths of an inch in diameter.

The formation of the I.D.T. requires only one vacuum deposition stage and, once the master mask defining the areas to be covered with metal is made, mass production of devices is possible.

Obviously the cost of such a procedure is minimal and complete standardisation is assured. This compatibility with micro-electronic techniques and promise of inherently economic production are major reasons for the present flurry of keen commercial interest.

VARIABLE DELAYS

A variable delay line is clearly now possible using a multiplicity of output I.D.T.'s to give varying delays; electronic switching being used to select the delay required.

It can easily be appreciated that just as a wave on the sea is damped a little by a ship riding on top of it, so an I.D.T. riding on a S.A.W. will reflect some of the incident acoustic energy, i.e. attenuate the ongoing wave. This can give rise to unwanted signals as they bounce between adjacent I.D.T's. Special techniques have been developed to overcome these problems.

The variable delay line mentioned will give only discrete delays and for a linearly variable delay two crystal substrates are used. From Fig. 5 it will be seen that if the lower substrate is held stationary while the upper one is moved mechanically, a continuously variable delay is possible.

An alternative to using longer and longer crystal substrates to obtain larger delays, and incidentally the larger the crystal the more difficult it is to obtain, is to use the helical delay line (Fig. 6).

This is so called because the delay path is a helix, the signal travelling round and round a specially prepared crystal many times. As can be seen from the diagram, transducers are placed along the S.A.W. path giving many temporally spaced outputs. These outputs can be up to several milliseconds after the original input pulse.

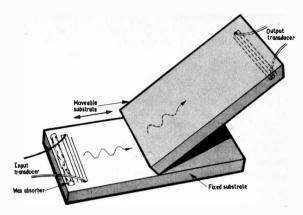


Fig. 5. A linearly variable delay fine can be created by having a moveable substrate in contact with a fixed substrate

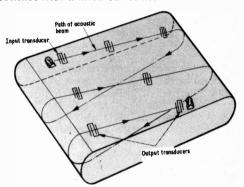


Fig. 6. For very long delays a specially prepared crystal substrate can be used to give a helical multiple-tap delay line

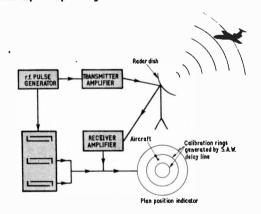


Fig. 7. A tapped delay line can be used in a radar system to give marker pips on a display screen

RADAR SYSTEMS

Applications for these types of delay lines include radar systems. Here the devices would provide range calibration and in featureless terrains simulate a background, i.e. clutter generation. Fig. 7 shows the simplified block diagram of such a radar range calibration system. A pulse generated by the r.f. pulse generator is transmitted via the radar dish to the atmosphere. If an object such as a plane is encountered a reflection of that pulse is returned to the dish some time later.

The sensitive receiver amplifier then generates a pulse on the plan position indicator (p.p.i.) the distance of which, from the centre of the screen, indicates the distance of the plane from the radar dish. The r.f. generator also passes a pulse simultaneous with the transmitted pulse, to the S.A.W. delay line. The delay line then gives delayed outputs which correspond, when displayed on the p.p.i., to specific distances from the radar dish. Thus range calibration of the radar system is achieved.

FILTERS

The bandwidth of the I.D.T. can be closely controlled as mentioned earlier by varying the number of fingers

Since the bandwidth is inversely proportional to the number of finger pairs, a filter, which allows only a limited range of frequencies through either side of its designed centre frequency, can be easily constructed.

The centre frequency is determined by the spacing between each finger, which is made equal to half the wavelength of the desired centre frequency. The resulting filter is of the bandpass type.

Filters with these characteristics are essential in every TV and radio, in the i.f. section for instance. The markets are obviously just right for a simple mass-produced device which requires no tuning up after fabrication. Enormous efforts have been turned in this direction and the complex requirements for a TV receiver are close to being obtained.

SIGNAL RECOGNITION

Suppose that we require a system whereby a plane flying around an airport control tower is able to tell the controller automatically that it wants to land. Let the plane have a transmitter which gives out a signal on a particular frequency.

The signal could be coded so that it is unique to that aircraft. A digital code would be suitable, i.e. a series of 1's and 0's, let these be modulated on the carrier as a bi-phase coding. This means that the carrier unchanged represents a 1 and a phase change of 180° of the carrier is a 0. If the code were four bits long, e.g. 1001 and two cycles of carrier are allotted to each bit, then the coded signal would be as shown in Fig. 8.

How can this signal of known form be recognised immediately it occurs notwithstanding the presence of much interference? Naturally S.A.W. come to our aid in the form of a tapped delay line or correlator.

The signal is first converted into a S.A.W. by the input I.D.T. As the signal feeds in it is compressed in length until it all lies along the substrate.

For simplicity the output I.D.T.s are made up of single finger pairs called taps. These taps will give a maximum output when a 1 or 0 bit of the S.A.W. appears beneath them, depending to which sum line the fingers are connected (Fig. 8). Thus if the taps are arranged 1001, as in the diagram, a S.A.W. of exactly that form at the correct frequency will cause them all to give a maximum output, simultaneously, when it appears beneath them.

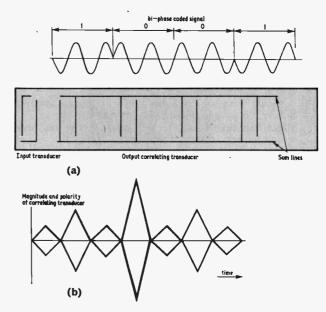
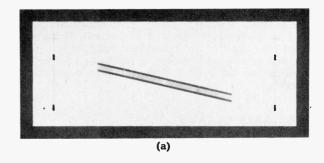


Fig. 8(a). A S.A.W. correlator may be used for detecting a particular sequence by arranging the taps to coincide with the desired signal. A typical output is shown in (b)



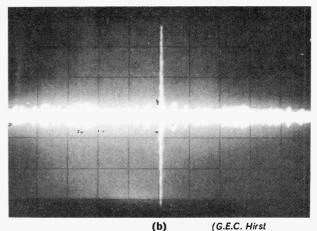


Fig. 9(a). Photograph of an actual S.A.W. correlator which is capable of recognising a particular 127 bit sequence. The output transducer is inclined to minimise distortion due to its length. Photograph (b) shows the large peak produced when the sequence is recognised

Research Centre)

The sum lines add these tap outputs to give a large pulse which indicates the signal has been recognised. Since the S.A.W. signal travels along the substrate under the taps it will cause small spurious signals before and after it is in the correct position to cause a large pulse; these are shown in Fig. 8b.

The photographs show a more ambitious example of a tapped delay line (Fig. 9a). It has 127 taps each

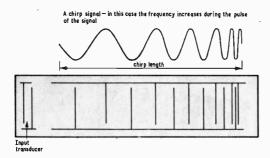


Fig. 10. By varying the spacing between taps as shown a "chirp" signal can be recognised

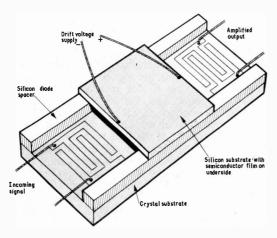


Fig. 11. The structure of an amplifier using S.A.W. principles

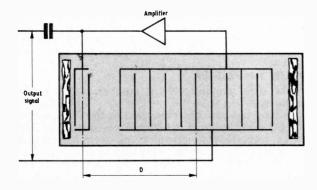


Fig. 12. A S.A.W. oscillator is produced by combining a bandpass filter with an external amplifier, the frequency of oscillation being determined by the spacing between taps on the S.A.W. crystal

of six finger pairs. Fig. 9b shows the output obtained from the large centre transducer, when the coded sequence is fed into the device. The detection pulse is easily recognised.

CHIRP DETECTION

If instead of bi-phase modulation a pulse of signal is used which is rising in frequency with duration, see Fig. 10. This could be detected by appropriately spacing the taps on a transducer as shown. The chirp system is common in radar systems.

AMPLIFIERS

The need for S.A.W. amplifiers is apparent in, say, a long delay line where attenuation of the wave by the crystal becomes the limiting factor in the achievable delay time.

The amplifier to be described makes use of the electric field generated by the S.A.W., travelling along a piezo-electric material, by causing it to interact with electrons travelling through this electric field.

Since the electric field is localised to the surface of the crystal, a conducting medium for the electrons has to be placed very near the surface, yet not touching, since this would distort the waves. Fig. 11 shows such an amplifier arrangement; a semiconductor film is used to conduct the electrons.

If the electrons are moving faster than the S.A.W. energy passes to the wave. Gains of up to 10⁷ times are being achieved at present. These require voltages of several kilovolts across the semi-conductor film.

It is interesting to note that if the electrons in the semi-conductor film are travelling slower than the S.A.W., energy is removed from the S.A.W. and the electrons benefit. This property is useful in absorbing unwanted signals.

OSCILLATORS

The oscillator structure (Fig. 12) combines the filter layout mentioned earlier with an external amplifier. This amplifier returns the signal taken from the filter output 1.D.T. to the input 1.D.T. with an excess of gain. The frequency of oscillation is determined by the output 1.D.T. and the spacing of the two transducers (D).

The number of finger pairs in the output I.D.T. determines the bandwidth whilst the spacing of the fingers sets the centre frequency. This bandpass filter then selects one of the many possible frequencies of oscillation, which are determined by the distance between the input and output I.D.T.s.

The range of frequencies presently possible is 20MHz to 1GHz. Oscillators above 300MHz will be small enough to fit inside transistor type TO-8 cans complete with their i.c. amplifiers.

Applications for these small, cheaply produced oscillators include TV tuners, low noise microwave sources and strain gauges.

SUMMING UP

It is seldom that one technique can achieve so much in such a short time. It is easy to see that this could be an area scheduled for intense activity and immense growth.

The combination of minimal power requirements (or none at all), microminiature construction, cheap mass production combined with a powerful signal processing capability must ensure the future of this branch of technology.



This amplifier has been designed on a value for money basis to give the highest standard of performance compatible with a small case and a components budget of about £30. The result is a circuit with an output of over 20 + 20 watts into 8 ohm loads at less than 0·12 per cent distortion, in a case measuring only 14in × 6in × 2in.

This has been made possible by the use of a toroidal mains transformer, by the small size of the latest electrolytic capacitors and by the use of the case of the amplifier as a heatsink.

PRE-AMPLIFIER

The circuit of the pre-amplifier is shown in Fig. 1.1. Here TR1 and TR2 form a complementary feedback pair. This arrangement has excellent bias stability due to the d.c. feedback through R12. Both transistors are low noise types and TR1 is run at a bias current of only 150µA to minimise noise.

The equalisation components are connected in the feedback loop. R14, R15, C9 and C10 provide equalisation for magnetic pickups to within ± 1dB of the R.I.A.A. curve, between 20Hz and 20kHz, whilst R13 and C8 give a flat frequency response for the tuner and auxiliary inputs. No special equalisation has been provided for ceramic pickups as these seem to be falling out of favour nowadays, but ceramic pickups can be used with the magnetic equalisation by connecting passive matching networks of the type shown in Fig. 1.3 inside the record player plinth.

The frequency response of the disc, tuner, and auxiliary inputs is shown graphically in Fig. 1.2.

Emitter follower TR3 provides a high input impedance for the tape input and also enables a tape A/B facility to be provided. This is of particular value with tape recorders having separate recording and playback circuits as it enables one to make a direct comparison between the signal source and the

recording. For example, if one wishes to make a recording of a radio programme the input selector is set to "Tuner" and the signal passes through the input stages and out to the recorder. The signal is recorded on tape and the tape playback signal appears at the emitter of TR3. By operating S2 one can then make a direct comparison between the input and the output of the tape recorder.

The tone control circuit is of the Baxendall type and uses an integrated circuit operational amplifier. The very high gain and large output voltage swing of this i.e. are advantageous in obtaining very low distortion and a good dynamic range, whilst the signal level in this stage is sufficiently high to make the noise negligible.

The characteristics of the bass and treble controls are shown in Fig. 1.4. With the tone controls flat the circuit has an overall gain of 2 and gives an output of 200mV.

The scratch filter is a second order type and gives an initial slope of 12dB per octave from its 3dB point at 5kHz. The response of the filter and its effect on the treble control is shown in Fig. 1.5.

MAIN AMPLIFIER

The main amplifier has a number of interesting features.

A long tail pair has been used at the input which increases the loop gain and reduces distortion, but more important it provides an accurate ground reference for the output. The d.c. potential on the output terminal will normally be less than 50mV and this will ensure that any d.c. current through the loudspeaker is of negligible proportions. However, if an output transformer, or a loudspeaker containing a matching transformer (such as the Quad electrostatic) is used the d.c. resistance is then very

^{*}Ferranti I td

SPECIFICATION.

Continuous Output Power

Load	Both channels driven	One channel driven
4 ohms 8 ohms	31+31 watts r.m.s. 23+23 watts r.m.s.	44 watts r.m.s. 30 watts r.m.s.
15 ohms	17+17 watts r.m.s.	19 watts r.m.s.

Measured at 1kHz

Toneburst Output Power

Load	Both channels driven	One channel driven
4 ohms	52+52 watts r.m.s.	57 watts r.m.s.
8 ohms	36+36 watts r.m.s.	42 watts r.m.s.
15 ohms	21+21 watts r.m.s.	22 watts r.m.s.

Measured with a 1kHz tone burst of 8 cycles on and 512 cycles off.

Distortion

15 ohm load—Less than 0·1 per cent at any power level up to 15 watts between 100Hz and 10kHz. Less than 0·02 per cent below I watt output.

8 ohm load—Less than 0·12 per cent at any power level up to 20 watts between 100Hz and 10kHz. Less than 0·02 per cent below I watt output.

4 ohm load—Less than 0·5 per cent at any power level up to 30 watts between 100Hz

and 10kHz. Less than 0.05 per cent

Frequency Response

Tuner and Aux. inputs $\begin{cases} -1 dB \text{ at } 28Hz \text{ and } 15kHz \\ -3 dB \text{ at } 17Hz \text{ and } 30kHz \end{cases}$ Tape input $\begin{cases} -1 dB \text{ at } 25Hz \text{ and } 30kHz \\ -3 dB \text{ at } 14Hz \text{ and } 60kHz \end{cases}$ Disc input—Within 1dB of the RIAA curve between

below I watt output.

Tone Control

Bass ± 12 dB at 100Hz, ± 18 dB at 30Hz Treble ± 12 dB at 10kHz, ± 16 dB at 20kHz

20Hz and 20kHz

Scratch Filter

- 3dB at 5kHz. Slope 12dB per octave

Inputs

Disc -3.5mV at 47k Ω RIAA equalised Tuner-100mV at 100k Ω Flat response Aux. -100mV at 100k Ω Flat response Tape -100mV at 100k Ω Flat response

Tape Output

100mV at 4·7k Ω Tape A/B facility

Signal to Noise Ratios

Unweighted figures measured with a bandwidth of 20kHz.

Weighted figures follow CCIR C curve. Volume control at max.

Tuner, Aux. Unweighted - 68dB,

Weighted -72dB

Disc. Unweighted -62dB,

Weighted -76dB
Tape Unweighted -76dB,

Weighted - 82dB

(Figures are relative to an output of 20 watts into 8 ohms)

MAIN AMPLIFIER ONLY Unweighted - 96dB (volume control at min.) Weighted - 100dB

Balance Control

Full rotation cuts off either channel

Dynamic Range

Disc input at IkHz = 32dB (i.e. input of I50mV)

Interchannel Crosstalk

-50dB

Stability

Unconditionally stable. Will drive electrostatic loudspeakers

Output Impedance

Less than 0-1 ohms

Dimensions

 $14 \times 6 \times 2$ in.

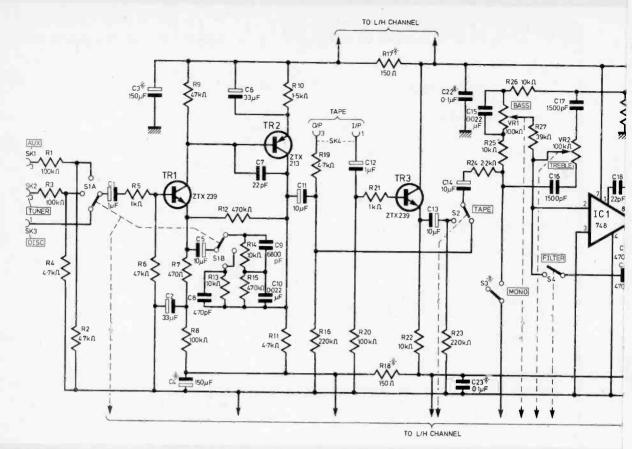
much lower than the speaker impedance and it is advisable to connect a resistor of about 0.5 ohms in series with the output of the amplifier.

This will reduce the output power by a few per cent but will minimise the d.c. current. In practice the resistance of the speaker leads will often be sufficient to provide the 0.5 ohm required.

The d.c. coupled output ensures that the speaker damping is maintained right down to d.c. giving a clean solid bass response. Normally, when a

speaker coupling capacitor is used, the reactance of this capacitor starts to become appreciable at low frequencies just when the most damping is required. The d.c. coupled output is made possible by the use of balanced positive and negative supply rails, and these also assist in obtainin clean symmetrical limiting under all load conditions.

The constant current source TR8 helps in obtaining low crossover distortion by providing a rapid transition of drive current between the two output



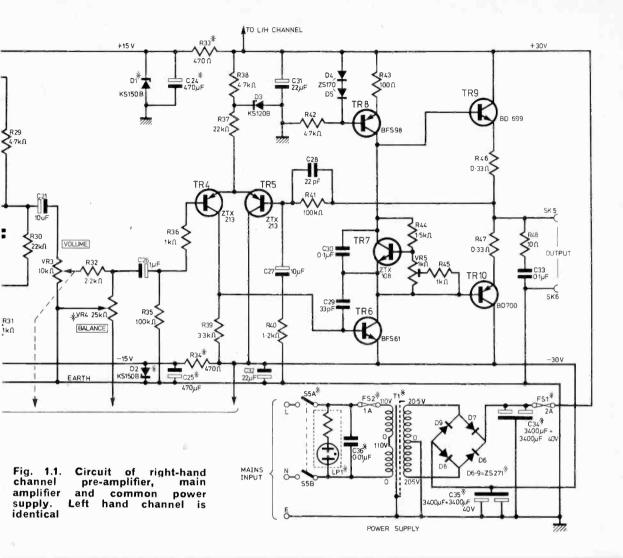
COMPONENTS . . .

Resistors		
R1, R101 100kΩ	R25, R125	10kΩ
R2, R102 4.7kΩ	R26, R126	10kΩ
R3, R103 100kΩ	R27, R127	39kΩ
R4, R104 4·7kΩ	R28, R128	4·7kΩ
R5, R105 1kΩ	R29, R129	4·7kΩ
R6, R106 47kΩ	R30, R130	22kΩ
R7, R107 470Ω	R31, R131	1kΩ
R8, R108 100kΩ	R32, R132	2·2kΩ
R9, R109 47kΩ	R33*	470Ω 1W
R10, R110 1.5kΩ	R34*	470Ω 1W
'R11, R111 4·7kΩ	R35, R135	
R12, R112 470kΩ	R36, R136	1kΩ
R13, R113 10kΩ	R37, R137	22kΩ
R14, R114 10kΩ	R38, R138	
R15, R115 470kΩ	R39, R139	
R16, R116 220kΩ	R40, R140	1·2kΩ
R17* 150Ω	R41, R141	100kΩ
R18* 150Ω	R42, R142	
R19, R119 4·7kΩ	R43, R143	
R20, R120 100kΩ	R44, R144	1·5kΩ
R21, R121 1kΩ	R45, R145	
R22, R122 10kΩ		0·33 Ω 2·5W
R23, R123 220kΩ		0·33Ω 2·5W
R24, R124 2·2kΩ	R48, R148	
All ½W 5% carbon film un	iless otherwi	se rated
Note: R17*, R18*, R33*	and R34" a	re common
		all other
components asterisked		

Potentiometers

VR1, VR101 $100 \mathrm{k}\Omega$ twin gang linear law (RS) VR2, VR102 $100 \mathrm{k}\Omega$ twin gang linear law (RS) VR3, VR103 $10 \mathrm{k}\Omega$ twin gang log law (RS) VR4* $25 \mathrm{k}\Omega$ single gang linear law (RS) VR5, VR105 $1 \mathrm{k}\Omega$ skeleton preset (RS)

Capacitors	
C1, C101	1μF 35V tantalum
C2, C102	33μF 16V elect.
C2, C102 C3*	150μF 16V elect.
C4*	150µF 16V elect.
C5, C105	10μF 25V tantalum
C6, C106	10μF 25V tantalum 33μF 16V elect. 22pF 160V polystyrene
C7, C107	22pF 160V polystyrene
C8, C108	470pF 160V polystyrene
C9, C109	6,800pF 400V polyester
C10, C110	022µF 160V polyester
C11, C111	10μF 25V tantalum
C12, C112	1μF 35V tantalum
C13, C113	10μF 25V tantalum
C14, C114	10µF 25V tantalum
C15, C115	·022μF 160V polyester
C16, C116	1,500pF 160V polystyrene
C17, C117	1,500pF 160V polystyrene
C18, C118	22pF 160V polystyrene
C19, C119	4,700pF 400V polyester
C20, C120	4,700pF 400V polyester
C21, C121	10μF 25V tantalum
C22*	0·1μF 30V disc
C23*	0·1μF 30V disc
C24*	470μF 25V elect.
C25*	470μF 25V elect.
C26, C126	1μF 35V tantalum
C27, C127	10uF 25V elect.
C28, C128	22pF 160V polystyrene
C29, C129	33pF 160V polystyrene
C30, C130	·1μF 250V polyester
C31, C131	22μF 63V elect.
C32, C132	22μF 63V elect.
C33, C133	·1μF 250V polyester
C34*	$3,400+3,400\mu\text{F} 40\text{V}$ elect.
C35*	3,400+3,400µF 40V elect.
C36*	-01μF 750V disc
	J. (J. 1. 1. 1. 1. 1. 1. 1. 1. 1. 1. 1. 1. 1.



ictors	
nmended type	Alternatives
ZTX239 Ferranti	ZTX384, ZTX109C
ZTX213 Ferranti	ZTX214, ZTX530.
	ZTX502
ZTX239 Ferranti	ZTX384, ZTX109C
ZTX213 Ferranti	ZTX214, ZTX502,
	ZTX531
ZTX213 Ferranti	ZTX214, ZTX502,
	ZTX531
BFS61 Ferranti	ZTX451
ZTX108 Ferranti	ZTX238, ZTX239,
	ZTX109
BFS98 Ferranti	ZTX551
BD699 Motorola	BD699A
0 BD700 Motorola	BD700A
3150B Ferranti	KS150A
S120B Ferranti	KS120A
170 Ferranti	Any ZS170/270
	series diode
170 Ferranti	Any ZS170/270
	mmended type ZTX239 Ferranti ZTX213 Ferranti ZTX213 Ferranti ZTX213 Ferranti ZTX213 Ferranti ZTX213 Ferranti BFS61 Ferranti ZTX108 Ferranti BFS98 Ferranti BD699 Motorola 0 BD700 Motorola 0150B Ferranti

series diode

ZS272, ZS274,

ZS276 or ZS278

Texas SN72748P

D6*, D7*, D8*, D9* ZS271 Ferranti

IC1, IC101 UA748CV Signetics

Switches

SIA, S1B, S101A, S101B 4-pole 3-way rotary (Lorlin)
S2, S102 2-pole changeover pushbutton (RS)
S3* 2-pole changeover pushbutton (RS)

S3* 2-pole changeover pushbutton (RS)
S4, S104 2-pole changeover pushbutton (RS)
D.p.s.t. rotary mains switch (RS)

Miscellaneous

T1 Gardners mains transformer type SL8, 20·5-0-20·5 volts

LP1 Neon panel lamp with internal resistor

Case—H.M. Electronics type GB3

Stereo jack socket, 4-DIN 5 way sockets, four 4mm sockets, five control knobs, three push-button switch buttons (RS), two Eagle 20mm fuseholders, 1A fuse, 2A fuse, four Lektrokit spring clips type LK2791 (1.5in), five way tagstrip. screws, spacers, grommets, aluminium angle, connecting wire.

A glass fibre printed circuit board printed with component locations, and a kit of semiconductor devices for this project are available from Davian Electronics, PO Box 38, Oldham, Lancs.

transistors. Bias transistor TR7 operates in the "amplified diode" mode and is thermally coupled to the output transistors by being clamped to the heatsink. This gives a great improvement in bias stability as any increase in heatsink temperature is compensated for by a reduction in bias.

OUTPUT TRANSISTORS

The output transistors TR9, TR10 on the circuit diagram are shown as single transistors for simplicity, but they are in fact monolithic Darlington pairs with a minimum current gain of 750 at 3 amps. These transistors have proven themselves to be electrically very robust and we have found that a 2 amp fuse in the positive rail is adequate to protect them against short circuits on the output.

Note that the fuse should be connected in the positive rail and not the negative rail. When the positive supply is removed the whole main amplifier is turned off because the bias is removed from the constant current source TR8, and also from the input stage TR4 and TR5. This in turn turns off

TR6.

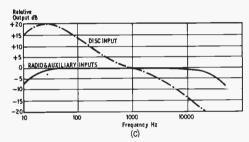
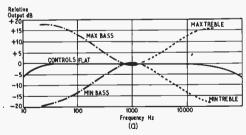


Fig. 1.2. Frequency response of disc, tuner and aux inputs



Frequency response of the tone Fig. 1.4. controls

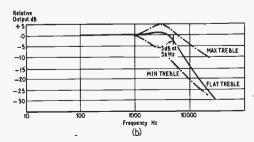


Fig. 1.5. Frequency response of the scratch filter

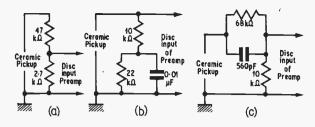


Fig. 1.3 (a). Circuit giving approximate matching for most types of ceramic pick-up; (b) circuit for matching the Decca Deram ceramic pick-up; (c) circuit for matching the Sonotone 9TAHC ceramic pick-up

POWER SUPPLIES

When one is designing a small, low cost amplifier the first refinement that one has to do without is the use of a stabilised power supply. This introduces a number of problems, but there are compensating

advantages.

The problems arise because of the lack of stabilisation. To get 20 watts output into an 8 ohm speaker we need a power supply which will deliver 45 volts on load. But when the amplifier is not giving any output and there is only a light load on the power supply, its voltage can easily rise to 60V. If we allow for mains voltage variations, then under the worst conditions the supply voltage could reach nearly 70V. With a stabilised power supply output transistors rated at 50V would have been satisfactory, but with our unregulated supply we need output transistors which will stand at least 70V. Not only this but all the electrolytic capacitors need to be conservatively rated as well.

The unregulated power supply does however have one very great advantage. A musical signal has a low average power level with occasional peaks of high power. For a short period (until the power supply voltage drops) an amplifier with an unregulated supply can deliver a power output very much greater than its continuous rating. This amplifier will deliver 23 watts per channel continuously into 8 ohms, but on a musical signal it is almost as

good as a 35 watt amplifier.

Lastly it might be as well to clear up exactly what we mean by continuous power. With 15 ohm loads the power dissipation is sufficiently low for the amplifier to be run at full sinewave power continuously. With 8 ohm loads it will also safely run continuously provided that it is placed in a well ventilated position where air can circulate freely around it, but the back of the amplifier tends to become rather hot after about 30 minutes of full sinewave power. With 4 ohm loads the amplifier should not be run at full sinewave power for more than about 10 minutes at a time, or the temperature of the output transistors may become excessive.

One does not normally listen to sinewaves of course and with a normal music or speech input the amplifier can be run continuously at full volume

without any reservations.

L

DISTORTION

At full output the distortion introduced by the other components in the hi-fi system will be much greater than that of any reasonable amplifier. Moving coil loudspeakers can generate up to 10 per cent distortion and even electrostatic types can give 0.5 per cent. A good modern f.m. tuner can generate 0.5 per cent., a tape recorder about 2 per cent, and a gramophone pickup can reach as much as 20 per cent on the inner grooves. Compared to these figures the performance of all but the most mediocre amplifiers is adequate at full output.

There lies the snag. All the signal sources may have considerable distortion at full output, but the distortion falls rapidly at lower levels. This is not necessarily the case with an amplifier. If crossover distortion is present the distortion may be only 0.1 per cent at full output but may easily rise to 1 per cent or more at low levels. Crossover distortion is particularly unpleasant because it generates high order harmonics, which are discordant and easily

perceived.

For crossover distortion to be negligible it should be less than 0.1 per cent at all power levels. Low order harmonic distortion is less objectionable and up to 0.5 per cent can be tolerated. So we can say that our amplifier should have a distortion specification of no worse than 0.5 per cent at full output, and below 0.1 per cent at all power levels below 1 watt.

The use of a constant current source in this amplifier has helped us to achieve a very low level of crossover distortion-typically about 0.01 per cent at 1 watt—and with 8 or 15 ohm loads the harmonic distortion is below about 0.1 per cent at all power levels up to full output. With 4 ohm loads the performance does not reach quite the same standard, but it is still below 0.1 per cent at 1 watt and 0.5 per cent at full output.

FREQUENCY RESPONSE

Many constructors are firmly convinced that a very extended frequency response is a good thing. This is a complete fallacy because-

- 1. Human hearing extends from about 20Hz to 20kHz at the best. There is some evidence that transients containing harmonic components above 20kHz can be distinguished but there is certainly nothing to be gained by extending the response past 40-50kHz.
- 2. There are very few loudspeakers with any useful response below 30Hz or above 20kHz.
- 3. There are no radio signals with any audio above 15kHz.
- 4. There are no records or cassettes with any audio above 20kHz.

In fact the only audio signal available which might have anything above 20kHz would be a very high quality tape recording of a live performance.

A very extended frequency response can be a very bad thing. If the low frequency response is very extended then low frequency noise from turntable rumble, warped or off centre records, or tape recordings can get through the system and cause the

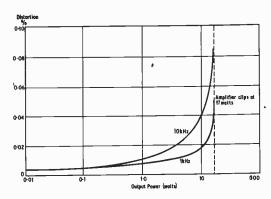


Fig. 1.6. Distortion against output power for 15 ohms measured at 1 and 10kHz

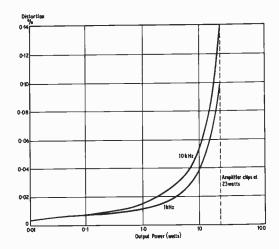


Fig. 1.7. Distortion against output power for 8 ohms measured at 1 and 10kHz

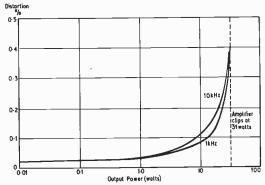
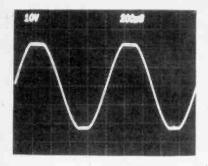


Fig. 1.8. Distortion against output power for 4 ohms measured at 1 and 10kHz

speaker cone to flutter violently. If the h.f. response is very extended then h.f. noise and multiplex sideband components can intermodulate with each other and with h.f. audio signals to produce audible distortion and noise. So what is the ideal response? Probably something like 20Hz to 50kHz

± 3dB.

This amplifier has been designed so that the frequency response falls rapidly below 10Hz, and so



2V 20068

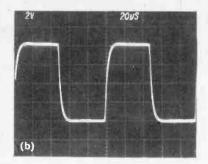
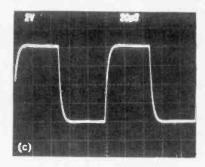
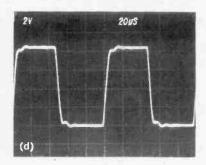


Fig. 1.9. Output waveform of the amplifier when slightly overdriven with a 1kHz sinewave showing the clean symmetrical limiting with freedom from latch-up.

Fig. 1.10 (a). 1kHz square wave response with 8Ω resistive load; (b) 10kHz square wave response with 8Ω resistive load; (c) 10kHz square wave response with load of 8Ω and $0.1\mu\mathrm{F}$; (d) 10kHz square wave response with a load of 8Ω and $2\mu\mathrm{F}$





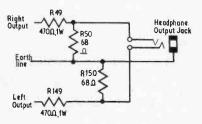


Fig. 1.11. Headphone attenuator circuit if required. Note that resistors have been omitted from Components List

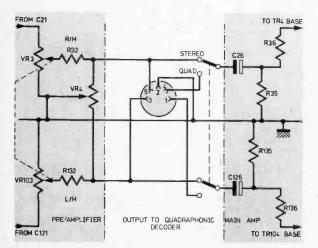


Fig. 1.12. Connections for quadraphonic decoder. DIN socket connections are as follows:—

1. Left channel front input from decoder

2. Earth

3. Left channel output to decoder

4. Right channel front input from decoder

5. Right channel output to decoder

a separate rumble filter is not necessary. The h.f. response of the tape input extends to 50kHz but the radio input has been restricted to 30kHz to attenuate multiplex and carrier components. With the scratch filter switched in all the inputs are 3dB down at 5kHz.

TRANSIENT RESPONSE

For good reproduction it is essential that the amplifier should have a good transient response with as little ringing as possible, even when fed into a highly reactive load such as an electrostatic loud-speaker.

In this amplifier a particular effort has been made to achieve a good transient response and Fig. 1.10 shows the performance of the amplifier under various load conditions. Note that with a $2\mu F$ capacitive load the ringing is of very low amplitude and soon dies away. If a 0.5 ohm resistor is connected in series with a $2\mu F$ load (as we recommended for the Quad speaker), then even this small amount of ringing is completely eradicated.

QUADRAPHONICS

No special provision has been made for quadraphonics in the prototype, mainly because of lack of space, but it is a simple matter for the enthusiast to adapt the circuit for use with a quadraphonic decoder.

All that is necessary is to break the connection between the preamplifier and the main amplifier and replace it with a switch and a DIN input socket, as shown in Fig. 1.12. The two front channels can then be fed back into the Orion main amplifiers whilst the back channels are fed to another amplifier. An additional pair of Orion main amplifiers would be ideal for this purpose.

Next month: Constructional details and setting up.

RE-READING one of the earliest publications on electronic music recently I came across the following statement made by Herbert Eimert, founder of the Cologne Radio Electronic Music Studio:

"That—electronic music cannot be performed on Instruments is due to the fact that the number of individual sound elements is so great that any attempt to find means of instrumental realisation is doomed to failure."

One's immediate reaction is to wonder whether he would have ventured to say this now, in the age of the synthesiser, when many pop groups have some kind of synthesiser.

Perhaps even twenty-odd years ago, when Eimert's article was written, there was little excuse for such a statement; the electronic and electrophonic instrumental field was by then quite sophisticated. A little relatively recent history may be pulled in here to support, or maybe excuse, his apparently negative remark.

The Years Between

In the years between the two world wars Arnold Schoenberg, Austrian composer, systematised the 12 notes of the traditional chromatic scale to produce music which did not rely on key (i.e. the predominance of one note over any other) but used all the available pitches equally. Amongst his pupils was Anton Webern who went a stage further than his master; rather than treating the 12 notes in a fairly conventional linear manner he perfected a style of writing which laid weight on each individual note as and when it occurred in the musical flow.

The dynamic levels, pitch and timbre were carefully controlled in his sparsely written aphoristic instrumental and vocal compositions.

Lionised Webern

After the war a group of young German musicians picked up the almost submerged threads of these revolutionary concepts and made Webern their idol. Reinforced by similar pre-organised, or serial, ideas put forward in works by the French composer Olivier Messiaen, these young intellectuals went on to produce a new music in which all the parameters available to music were fully exploited, almost mathematically. The musical results were often impossible to perform by human beings, yet the European scene rapidly became thick with "avant-garde" concerts, many of dubious integrity.

That any worthwhile music has survived this period is a minor miracle, given the arrogance of the exponents and the loud raspberries



of the musical press.

It was during this babel of activity that the tape recorder came into general use, and with it the very latest in sensational sound, electronic music. Anything essentially of an acoustic nature was taboo. Needless to say electronic compositions were meticulously edited and fixed for all time on magnetic tape.

Colouristic Effect

Meanwhile in France there was another little sonic revolution taking place. Whilst those who whored after strange gods studied the latest serial techniques in Germany, others stayed behind to practice the art of colouristic effect for which the French have always been renowned.

Pierre Schaeffer founded a studio in Paris dedicated to the study of the physics and psycho-physics of acoustic phenomena and to the production, on discs initially, of musical compositions which took raw, natural sounds as their starting point. Plain aural effect was the ambition; Schaeffer and his colleague Pierre Henry had little time for the intricate number games which the Germans were playing. The search for "musique concrete" [natural or non-abstract music] ended with the arrival of the tape recorder and the simplicity of editing magnetic tape.

Partisan Factions

So electronic music began lite in two partisan factions: the German-influenced found the French "musique concrete" positively naive and artless, whilst the French considered the Germans inhumanly mathematical

and equally artless. Both came together, however, in considering their respective tape compositions to be one-off, once-and-for-all performances.

Live electronics

So Eimert was right. Given the complexity of a totally serial composition with its rapidly changing rhythmic, pitch, spatial, dynamic and timbre elements, only a taped sequence could do justice to a particular concept. No amount of juggling around with electronic organs and peripheral sound effect units would reproduce on the concert platform what could be realised on tape.

The same applies to Schaeffer's collages, where natural sounds, recorded wherever they existed, were processed electronically, the result defying any musical instruments—save perhaps the Melotron, which is nothing more than a filing cabinet of tape recordings with a piano keyboard attached. Eimert lacks credence in his underlying assumption that Electronic Music is a style rather than a medium of expression.

Today's Electronic Music

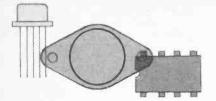
In today's Electronic Music the factions have largely disappeared; Stockhausen has cast off the strait-jacket of total serialisation and in his taped works uses the human voice, gramophone records, short-wave radios and acoustic instruments. Schaeffer and his colleagues rely quite heavily on purely electronic sound producers.

Along with this fusion of two differing approaches to taped material came the feeling that performers ought to take some share in the interpretation of this kind of music; hence "Live electronics".

Pop Sounds

By far the largest purveyors of live electronics are the pop groups. (Only a pedant would argue that their music is not strictly electronic.) Electronic instruments work admirably here since most of the music performed is geared to the traditional pattern of twelve notes to the octave and is old fashioned enough to use fairly common juxtapositions of these notes. Above all, pop-music is primarily melodic music and this means that a given instrumental line is unlikely to require rapid changes of timbre.

To some extent "serious" live electronic groups have taken a lead from the pop world; the music they perform is no more complex to realise than an equivalent piece of acoustic instrumental music. Should they require complex arrangements of sound these still have to be prerecorded on tape.



THE NEXT DECADE?

Final selection of readers' predictions

ARMCHAIR VIEWING

BELIEVE that in the next decade we could well see many changes in our everyday life due to electronics. Instead of travelling to conferences, businessmen might stay at home and "attend" the conference via video and audio links to a central conference control, through which they would be able to see and hear their colleagues. They would also be able to "look up" relevant information through the conference control and have it displayed on their screens for as long as they wish.

We could watch any television (if it could rightly be called this) programme

whenever we pleased.

The printing world can throw away its presses, for we will be able to dial for, say, the front page of today's newspaper and have it displayed on our "newscreens." Yes! even our beloved Practical Electronics will come to us like this. (Don't worry Ed., you will still be needed.)

No-one will have to think as much, for we will all have our own computer calculators, and musical instruments will be replaced by synthesisers, so that the Pablo Casals' of tomorrow will play upon streamlined keyboards instead of peculiarly shaped pieces of wood.

All this sounds rather unlikely? Well, you wait and see.

S. J. Baxendale.

COMMON PLACE

N the next ten years no progression is forseeable in the direction of component miniaturisation, owing to the impossibility of decreasing pin and encapsulation size. It is likely that more will be fitted into a single module—Mr. Shaw's synthesiser on a chip, for example.

Specialisation of these circuits is almost inevitable, specifications going further and further towards the extreme, with fantastic power handling and even more fantastic frequency ranges, now only the dream of hi-fi enthusiasts. Technology cannot be underrated in sorting out the problems which may make these two characteristics (now) incompatible.

For the home constructor, today's high-powered technology may be to-morrow's workbench experiment. A home made integrated circuit kit for example; which is not such a far fetched idea if one looks at the basic simplicity

of overlaying the semiconducting layers, assuming there will be enough semiconducting material left by tomorrow!

Soon one may be able to leaf through a catalogue of "surplus" equipment, which at present is described as a technological breakthrough. Basic techniques could soon be outmoded, soldering and the printed circuit board might soon be replaced by the plug-in module.

In conclusion, it could be said that in the next decade, though there may not be the advances of the like of the integrated circuit in the last ten years, technology will push its usage to the limits. Also, present technology will be commonplace to the constructor in not so many years.

N. J. Eastaugh

creations, having purchased programmes from the lists in the advertisement pages of his favourite journal, P.E.

Seated at the console of his electronic organ, engaged in extempore musical invention, such will be the interaction between man and his electronics that the player may well exclaim from the profundity of philosophical doubt, "Just who is playing this machine?" The folly of the decade will be an attachment for colour TV receivers to permit the display of electronically simulated goldfish.

D. Letts

MAN AND HIS ELECTRONICS

THE philosophy of electronics in the next decade is summed up by "digital is best, and smaller is better".

Compromised by cost, and using the results of researches into the nature of human perception, manufacturers will set lower standards of sound and image reproduction.

Enthusiasts will reconstruct the electronic achievements of the 1920's with antique components or replicas made by new cottage industries.

A wide range of games-machines designs will be published for the constructor. At best, they will be war games, or a form of Monopoly in which the players are relieved of the arithmetic of accounting and cannot break the rules.

The power of present-day minicomputers will be available in a single i.c. costing £20, or less.

The military applications of microelectronics will shock and horrify us, but the constructor will find light relief from the problems of the day by assembling the parts of a micro-computerised mouse for his cat to play with.

There will be great interest in the generation of special effects. Today's Wind and Rain generator will be supplemented by tomorrow's Thunder and Lightning, the degree of authenticity rising according to ingenuity and the size of purse.

Electronic musical instruments will have substantial computer power, and the constructor will spend more time in programming than constructing his

HIGHER QUALITY

THE next decade will not unveil any major electronic breakthroughs but instead, in a time of financial instability, manufacturers will concentrate on improving the quality of goods already available, the risks being too great for a large scale venture into something radically new. Maximum profit being essential, the fear of failure in a component would be too daunting.

Transistors and integrated circuits will continue to be the main "workbench" of electronics, both on an amateur and professional scale. The valve will still command itself a place but will always run a poor second to the modern semiconductor, as I feel it does now.

With everybody striving to attract the prospective customer to their particular line of components there will inevitably be many new i.c.s on the market, having better power handling capacities and such-like in an effort to better their predecessors, but none of them will be fundamentally different.

A shortage of raw materials, evident at the present time, will mean price rises and delivery delays (nothing new!) This could also result in the amalgamation of many small component retailers who would otherwise be forced into liquidation by the bigger concerns cornering all the custom.

The overall trend will be to encapsulate the components in plastic containers which constantly seem to diminish in size

Finally, the high degree of competitiveness between manufacturers will result in a higher quality of components—which can only be to the good of the constructor in the end.

M. S. Johnson

ELECTRONIC LABEL

THE present trend of improvement in electronic technology is likely to provide some new and perhaps exciting uses of the circuits as we know them. It may be taken for granted that they will reduce in size, become more reliable and presentable, and who can say how long they would last.

The reading of domestic meters by remote computer, preparation of account, and even automatic payment of the bill by pre-arrangement, is but one aspect.

It is likely that television receivers will be used in the dual role of entertainment or access to information, such as availability of goods or prices. The telephone line perhaps, would be switchable to a number of facilities, with a "back up" visual display.

One possible change is in the use of pocket, or desk calculators. It is quite feasible to envisage these being used as mini computers without major circuit change. For instance, with a microfilm attachment, and "access code" operation using the same styled keypad, the film could be rotated by the command word, and illuminated on a slightly enlarged screen.

Finally, how about the "Electronic Label", a microcircuit, so small that it could be incorporated into almost any article at the manufacturing stage, and with its own identity calling signal, quiet, unless called by a master beacon. Nasty though, if you want to go into a pub for a quick one and the wife is looking for you!

A. J. Williams

DIGITAL TRACKS

THE next decade will be dominated by political change, which will include higher taxation for the individual, and wage rates so high that the cost of maintenance and repair will often exceed the value of the article to be repaired.

The first item will make it unlikely that the colour television boom will be repeated with video cassette recording or the Ceefax/Oracle system. However, this may well give rise to a new system of video recording towards the end of the period using digital methods with about 150 narrow tracks instead of the helical scan system with its mechanical complexity.

Such a change would be less likely if a large number of helical scan recorders were already in use.

The lack of money in the hands of individuals will be compensated by more in the hands of government, who will spend more on electronics.

One possibility taken at random could be a distress calling system for old and disabled people. The caller presses a button that activates a device to send a high frequency signal into the local mains electricity network. This signal is modulated by a series of pulses representing a number allocated to the caller.

A warden who has a receiver on the same local mains ring is alerted that the caller is in trouble.

A second is the introduction of automated speed traps, made necessary by increasingly restrictive speed limits due to overcrowding and fuel shortages. These would photograph offenders' cars showing the speed on a print-out. An official would collect the photographs every so often. To avoid the possibility of malfunction causing injustices, offenders would be prosecuted if caught more than once in a set period of time.

The second item, wages, will lead to novel methods of improving reliability of electronic goods. Touch-operated integrated circuit sub-systems will replace potentiometers and switches, and possibly i.c. optical modulators and demodulators coupled with optical fibres will replace plugs and sockets for connecting audio discrete units.

J. de. Rivaz

INTEGRATED I.C.'s

TEN YEARS is not a long time, but I believe that man will advance more in the next ten years than in the last ten. After long consideration I came to the conclusion man's advancement was of a logarithmic nature and not a linear one. Our greatest advancement will be in the field of i.c.s and their interconnection.

Soon we will see the inductor being incorporated in i.c.s and midway through the decade we will see the key to a new kind of electronics. No more will i.c.s be coupled by wires, coupling will be similar to the lecher lines used in u.h.f. tuners. Simply placing the i.c.s one on top of another, rather like the child's building bricks only in a miniature form. From this giant leap electronics will virtually know no bounds. The solid state power pack recharging itself from air, light, heat, or vibration, for example.

Air and light may be used, as in plants, to create chemical changes. These we can change into electrical energy. Heat given off due to power loss will be channelled back to the power pack to help recharge power cells. Vibration as used in self winding watches, used to generate power.

These advancements will lead to vast developments, such as the pocket computer, rather similar to the pocket calculator we have today, only thousands of times better and incorporating an audio output.

Programme cartridges will also be solid state, as in microfilm one small block containing many hundreds of hours of information.

Such an instrument would place the home constructor almost on a par with the professional.

Roll on 1984.

A. Tannock.

DIODES AND ALL

EXPERIENCE shows that the development of technology follows something akin to a log. law curve. In semiconductor technology there has been a veritable explosion of new ideas and techniques. If interpolation of this curve is attempted then the results can be somewhat surprising.

Today's computers using holographic or ferrite memory stores are far too cumbersome, slow and expensive. Development of heavy metal organo compounds already well under way should bring about the production of high temperature (in excess of 100°K) superconducting memory stores:—possibly a very primitive forbear of Asimov's positronic brain!

For constructors, l.s.i. circuits are already available, the question remains how large (or small) can they get and for what purpose. Pocket calculators already have chips containing thousands of active elements, and I rather think that today's constructor in ten years will be in much the same boat as those are now who lament the passing of the valve, we shall be lamenting the passing of the discrete transistor.

Looking back, most developments seem to stem from the humble diode, the latest being the ubiquitous l.e.d. so the logical development of this would be the "light emitting transistor"—alter the base bias and "hey presto"; modulated light output.

E. J. Marchant.

SIMPLIFICATION

THE following decade, for myself and fellow constructors, should bring simplification in the form of reduced wiring and soldering for more complicated circuits with increasing use of i.c.s.

With spiralling cost of most products, the hope of continued amateur construction may lie in continued progress in i.c. technology to reduce manufacturing costs.

For the home constructor whose use of i.c.s was exclusively bipolar (TTL), he may find himself making adjustments in the not too distant future towards a different form (CMOS). In this increasing energy-conscious world, the life of the bipolar form of logic could be drawing to an end, with less power consuming logic forms such as CMOS becoming more and more popular.

Amateurs and professionals over the coming years will be made more power conscious and a tendency towards battery supplies in contrast to more expensive a.c./d.c. Transformation will be encouraged as a practical step for the home constructor.

Manufacturers will be responsible for this trend, as they become more aware of producing the form of logic which will be more financially secure for the future.

S. NaismIth

Now-two fascinating ways to enjoy saving money!

NEW! Sinclair Scientific kit \$19.95

Britain's most original calculator now in kit form

The Sinclair Scientific is an altogether remarkable calculator.

It offers logs, trig, and true scientific notation over a 200-decade range – features normally found only on calculators costing around £100 or more.

Yet even-ready-built, the Sinclair Scientific costs a mere £32.35 (including VAT).

And as a kit it costs under £20!

Forget slide rules and four-figure tables!

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sin and arcsin,

cos and arccos,

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 log_{10} , antilog₁₀, giving quick access to x^{Y} (including square and other roots),

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In fact, virtually all complex scientific or mathematical calculations can be handled with ease.

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No. Powerful though it is, the Sinclair Scientific is a model of tidy engineering.

All parts are supplied – all you need provide is a soldering iron and a pair of cutters. Complete step-by-step instructions are provided, and our Service Department will back you throughout if you've any queries or problems.

Of course, we'll happily supply the Scientific or the Cambridge already built, if you prefer — they're still exceptional value. Use the order form.

Components for Scientific Kit (illustrated)

- 1. Coil
- 2. LSI chip
- 3. Interface chips
- Case mouldings, with buttons, windows and light-up display in position
- 5. Printed circuit board
- 6. Keyboard panel
- 7. Electronic components pack (diodes, resistors, capacitors, etc)
- 8. Battery assembly and on/off switch
- 9. Soft carrying wallet
- 10. Comprehensive instructions for use

Assembly time is about 3 hours.



Features of the Sinclair Scientific

12 functions on simple keyboard
Basic logs and trig functions (and their inverses), all from a keyboard as simple as a normal arithmetic calculator's. 'Upper and lower case' operation means basic arithmetic keys each have two extra functions.

Scientific notation Display shows 5-digit mantissa, 2-digit exponent, both signable.

200-decade range 10⁻⁹⁹ to 10⁺⁹⁹.

Reverse Polish logic Post-fixed operators allow chain calculations of unlimited length – eliminate need for an = button.

25-hour battery life
4 AAA manganese alkaline batteries (e.g.
MN2400) give 25
hours continuous
use. Complete
independence
from external
power.

Genuinely pocketable 41/3" x 2" x 11/16". Weight 4 oz. Attractively styled in grey, blue and white.



Sinclair Cambridge kit 3



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In less than a year, the Cambridge has become Britain's most popular pocket calculator.

It's not surprising. Check the features below – then ask yourself what other pocket calculator offers such a powerful package at such a reasonable price.

Components for Cambridge Kit

- 1. Coil
- 2. LSI chip
- 3. Interface chip
- 4. Thick film resistor pack
- Case mouldings, with buttons, window and light-up display in position
- 6. Printed circuit board
- 7. Keyboard panel
- 8. Electronic components pack (diodes, resistors, capacitors, transistor)
- 9. Battery clips and on/off switch
- 10. Soft wallet

Assembly time is about 3 hours.

Scientific

Price in kit form £19.95 inc. VAT. Price built £32.35 inc. VAT.

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The Sinclair Cambridge and Scientific

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All parts are tested and checked before

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form below and slip it in the post today.

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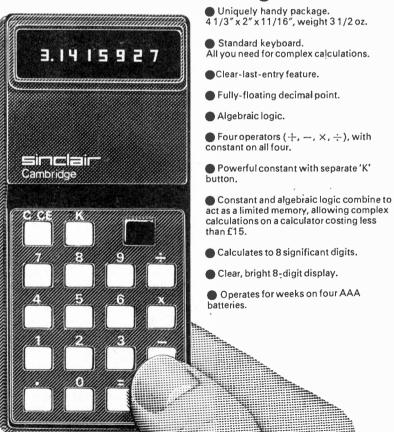
Simply fill in the preferential order

your money without question.

Cambridge

Price in kit form £14.95 inc. VAT. Price built £21.55 inc. VAT.

Features of the Sinclair Cambridge



To: Sinclair Radionics Ltd. FREEPOST St Ives, Huntingdon, Cambs. PE174BR Please send me Sinclair Scientific kit at £19.95 Sinclair Scientific built at £32.35 Sinclair Cambridge kit at £14.95 Sinclair Cambridge built at £21.55 All prices include 8% VAT. *I enclose a cheque for £....., made out to Sinclair Radionics Ltd, and crossed. *Please debit my *Barclaycard/ Access account. Account number *Delete as required. Signed Name Address Please print, FREEPOST - no stamp needed. PE/1/75



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PE MINISONIC

By D. SHAW

PART THREE

- Keyboard Controller
- Noise Generator
- Ring Modulator
- Power Amplifiers

THIS month the remainder of the electronic circuitry in the synthesiser is described which includes KEYBOARD CONTROLLER, RING MODULATOR, NOISE GENERATOR and POWER AMPLIFIERS.

THE KEYBOARD CONTROLLER

The KEYBOARD CONTROLLER as illustrated in Fig. 3.1 is a relatively simple means of providing a range of voltages which, when applied to the input of a vco, cause it to oscillate over a range of pitches normally associated with a chromatic scale or, alternatively, over a range of pitches quite outside what might be termed normal musical acceptance.

IC1 and IC2 are inverting operational amplifiers whose outputs are linked by a chain of resistors the junctions between which are connected to the keyboard contacts. R5 and VR1 form a divider between the positive rail and ground such that the swing of the potentiometer covers a range of about 4.7 volts.

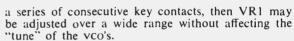
The wiper of VR1 is linked to both i.c.'s so that the output of these devices will track, in unison, the setting of VR1. R1 and VR2 form a second divider between the negative rail and ground with the wiper linked to IC1 only. Thus VR2 is able to provide an offset to IC1 which is variable over 4.5 volts.

The purpose of the voltage difference between the swings of the two potentiometers, is so that, under normal conditions, the key contact voltages can never go positive and thus drive the VCO's into saturation.

SPAN AND TUNE CONTROLS

The KEYBOARD CONTROLLER can be matched to a wide range of keyboard sizes and vco control voltages.

If, for example, a two octave keyboard is to be used and the required control voltage for the vCo's is 600mV per octave, then VR2 (the "Span" control) will require to be offset by 1.2V with respect to the inverted value of VR1's setting. Once this has been used and the required control voltage for the vCo's are able to reproduce a chromatic octave by making



In simple terms the "position" of the two-octave keyboard may be varied over the audio frequency range and the "white" notes may be made to play in any required key signature. This latter feature will commend itself to those "play-it-by-ear" musicians who may sometimes find difficulty in translating a well known melody in the key of C into its correct signature.

For more serious applications, however, the ability to swing the keyboard "position" enables the Minisonic to play in tune with a number of conventional acoustic instruments which may, themselves, not be precisely "spot-on" as far as tuning is concerned.

KEYBOARD RESISTOR CHAIN

No setting-up is required for the KEYBOARD CONTROLLER other than to check that the outputs of both IC1 and IC2 respond correctly to the settings of VR1 and VR2. Fig. 3.1 gives a table of resistor values which may be used for the divider system on keyboards of various sizes.

It will be noted that the overall value of resistance in each case is approximately the same in order that the loading on the i.c.'s will vary by a minimum amount regardless of the size of keyboard employed.

THE "HOLD" OR ANALOGUE MEMORY

Although covered by the general heading of KEYBOARD CONTROLLER the HOLD circuit is a quite separate entity which fulfils an important function in the scheme of the synthesiser.

Last month it was indicated that the ENVELOPE SHAPER could give a decay characteristic lasting up to 16 seconds. In other words, from the instant the key contact is broken, the audio signal will continue—at a diminishing level—for the prescribed period. It is obvious therefore that, for the best effect to be achieved, the VCO frequency must remain constant for the period over which the decay is taking place.

With the key contact broken so too is the VCO programming voltage disconnected unless there is some means by which the VCO can continue to be programmed regardless of key contact condition. The HOLD circuit provides the means whereby the VCO can continue to oscillate at the frequency prescribed by the last programmed voltage either until the ENVELOPE SHAPER completes its cycle or until another voltage is programmed in.

HOLD CIRCUIT

The circuit of the HOLD facility is shown in Fig. 3.2a. IC3 is an operational amplifier in which the output signal is divided by means of VR4, R8 and R9 to provide balanced levels of positive and negative feedback.

When the balancing is carefully done the circuit is theoretically capable of presenting an infinite impedance to incoming signals. In practice, however, it is more usual to calculate the input impedance on the basis of the parallel value of the feedback resistors times the open loop gain of the amplifier. Thus the input impedance is of the order of 2,500 megohms.

The hold capacitor (C2) is, ideally, a low leakage type. A charge applied to C2 is reflected at the output of IC3 with any drift at the output due to a combination of capacitor leakage and minor thermal effects within the i.c.

COMPONENTS . . .

KEYBOARD CONTROLLER

Resistors

R1 10kΩ

R2-R4 47kΩ (3 off)

R5 9·1kΩ

R6, R7 47kΩ (2 off)

R8, R9 20kΩ (2 off)

R10 $47k\Omega$

R11 et seq See text

Potentiométers

VR1, VR2 10kΩ linear carbon (2 off)

VR3 $10k\Omega$ sub-miniature

skeleton preset

horizontal

VR4 10kΩ 15-turn preset

Capacitors

C1 1,000pF

C2 1µF 63V polycarbonate

Integrated Circuits

IC1-IC3 Type 741 8-pin d.i.l. (3 off)

Miscellaneous

SK1 2mm socket

KEYBOARD CONTROLLER

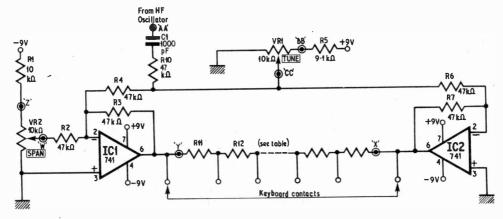


Fig. 3.1. Circuit of the KEYBOARD CONTROLLER (excluding the HOLD circuit). The table (below) shows values of resistors (R11 et seq) and numbers required for various length keyboards. This applies to both printed circuit and conventional keyboards

KBD DIVIDER RESISTORS Resistor Number off Size 150Ω 13 1 octave 82Ω 25 2 octave 51 or 56Ω 37 3 octave 39 or 43 Ω 49 4 octave 33 or 36 Ω 61 5 octave

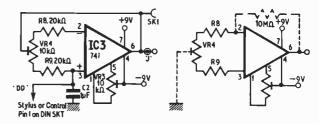


Fig. 3.2(a). The circuit of the HOLD section of the KEYBOARD CONTROLLER. (b) It is important that this circuit should be adopted when nulling the HOLD offset. Temporary links are shown dashed. The feedback resistor should be 10M Ω or more

It is possible to balance the circuit such that the output drift is better than ImV/sec but to do so requires considerable patience and care, particularly when nulling the offset. The circuit for this latter procedure is shown in Fig. 3.2b. The component assembly should be as shown on the circuit board layout but the wiper of VR4, instead of being linked direct to the output of IC3, is temporarily connected to the 0V rail.

A second temporary feature is the inclusion of a high value feedback resistor (ideally 10MΩ or more)

as shown hatched in Fig. 3.2b.

Adjust VR4 so that its wiper is close to the centre of travel and, with power on, adjust VR3 until the output of IC3 is *precisely* zero volts. The temporary links and feedback resistor may now be removed and the circuit completed as shown in Fig. 3.2a.

Minimising the drift in the HOLD circuit is best done by ear, i.e. using the Minisonic vco's rather than an oscilloscope as part of the test equipment. Details of this procedure will be included as part of

the final setting up.

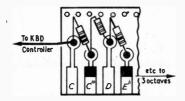


Fig. 3.3(a). Wiring of the edge connector strip as used on the prototype. Resistors are wired in from the conductor side of the board. Excess wire on the other side of the p.c.b. should be trimmed off and filed flush so that the board may be glued to the front panel

THE KEYBOARD

The Minisonic offers the possibility of being operated with a number of keyboard options, the cheapest being the edge-connector type. Other

options will be discussed next month.

A printed circuit keyboard was adopted in order that the instrument could be both compact and fully self-contained. In the prototype a three-octave keyboard was made up using a standard edge-connector strip as shown in Fig. 3.3a but satisfactory operation could only be achieved after much practice due to the narrow conductors involved. Mounting of the divider resistors should be generally as shown in the diagram.

COMPONENTS . . .

HARDWARE

Control Knobs, 14mm—19 with skirt, 1 without skirt (ReAn Products—see *Market Place*) On/off d.p.d.t. toggle

Battery connectors, PP9 2-off positive and negative

0-1in matrix Veroboard 45 × 34 holes

Keyboard 37-way edge connector strip (or see text)

Materials for stylus (see text) 5-pin 180° DIN socket

STYLUS

In the first prototype the stylus employed two contacts and was illustrated on the front cover of the November issue. The double contact, however, greatly added to the difficulties of playing the instrument and thus modifications were carried out so that a single contact stylus could be employed.

Perhaps the simplest stylus involves the adaptation of a ball-point pen (see Fig. 3.3b). If this method is chosen it is important that all traces of ink are removed from the ball end using an organic solvent before any attempt is made to solder in the wire lead.

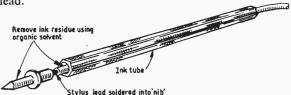


Fig. 3.3(b). A suggested construction method for the stylus using an old ball-point pen

Note that organic solvents should be treated with caution since most of them give off a vapour which can be harmful if inhaled continuously. The assembly when completed should be potted within the lower half of the pen by means of Araldite or Silicon Rubber Compound.

Those constructors having access to a lathe could make up a stylus from a piece of ‡in brass rod. If this method is used it is important that the extreme tip of the stylus should be rounded off and well

polished to ensure a good contact.

ULTRASONIC TRIGGER SYSTEM

(The circuitry in this section is the subject of a Patent Application)

The changeover from a double-contact to a single-contact stylus presented a difficult problem simply because the signals required to set the HOLD circuit and to trigger the ENVELOPE SHAPERS are essentially incompatible. Direct coupling between the inputs of these two circuits was therefore not possible since, once the HOLD capacitor was charged, the d.c. level would remain on the stylus lead and the ENVELOPE SHAPER in the "on" condition, until the charge on the HOLD capacitor had leaked away.

This would occur quite rapidly in the circumstances thereby giving rise to an undesired portamento effect. Similarly it was not possible to decouple the ENVELOPE SHAPER from the stylus lead since so doing would restrict the "attack" phase to one rate only—and that very fast. The solution proved to be the application of a principle which is believed to be unique in electronic musical

instruments.

HF OSCILLATOR

A HIGH FREQUENCY OSCILLATOR is coupled directly into the KBD CONTROLLER in such a way as to distribute the signal evenly across the divider. The stylus lead which now goes direct to the HOLD capacitor is also connected through a decoupling capacitor to an a.c. detector circuit which, through an integral switch, is used to trigger the ENVELOPE SHAPER.

Four components only go to make up the HF

OSCILLATOR which is shown in Fig. 3.4.

VR1 controls the frequency of operation by prescribing the proportion of positive feedback and thereby varying the peak to peak value of the output signal. With the component values given the frequency range is from 2kHz at 18V peak-to-peak, to 250kHz at 80mV peak-to-peak. Output waveforms are also shown in Fig. 3.4.

OSCILLATOR FREQUENCY

The optimum setting of the HF OSCILLATOR is 40kHz at 6V p-p as measured at point "AA." The attenuating effect of C1, R10 and VR1 in the KBD CONTROLLER will combine to reduce the signal to 500mV p-p measured on the keyboard contacts.

It should be noted however that the setting of VR1 in the controller will affect the level of the h.f. signal—the lower the setting of VR1 the lower will be the level of the signal on the contacts. This is not really a problem since the detector sensitivity is around 50mV and also, for most applications, it will be found that VR1 will require to be at a relatively high setting.

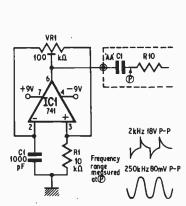
HF DETECTOR

The circuit of the DETECTOR is shown in Fig. 3.5. IC2 is a high gain follower decoupled from the stylus lead by means of C2. C1 provides additional decoupling for the stylus lead thereby ensuring that hum signals which may be included in the lead do not cause triggering of the envelope shapers. C4 and C5 provide frequency compensation for IC2 which is a 709 operational amplifier to give the advantage of the higher gain bandwidth offered by this device.

The output of IC2 provides drive to TR1 the collector of which is coupled through R6 to the bases of TR1 and TR2 on both envelope shapers. (Note that this latter coupling is via the DIN socket and JK1 on both envelope shapers.) C6 blocks any d.c. appearing at the output of IC2 while R5 sets a current limit.

Under quiescent conditions the output of IC2 is nominally zero volts and TR1 is off. An a.c. signal of sufficient level on the stylus lead will cause IC2 to follow and each positive excursion of IC2 output will switch TR1 on causing the collector to go to about —8.5 volts. The ENVELOPE SHAPERS thus start to attack and C7 receives a negative charge.

H.F. OSCILLATOR AND DETECTOR



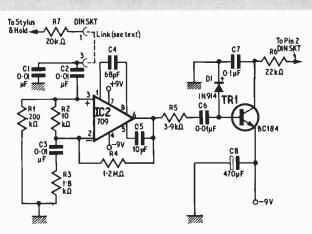


Fig. 3.4. Circuit of the HF OSCILLATOR. Components in the dotted box are on KBD CONTROLLER circuit and are mounted on main board. Typical waveforms at different settings of the VR1 are also shown

Fig. 3.5. Circuit of the HF DETECTOR. Resistor R7 is for isolation and was $20k\Omega$ in the prototype. The DIN socket is for external keyboard attachment and wiring options will be described next month. C1 is mounted on the DIN socket

HF DETECTOR COMPONENTS. Resistors R1 200kΩ 3.9kΩ HF OSCILLATOR R5 R2 10kΩ 22kΩ R6 Resistors R3 1.8kΩ 20kΩ (see text) $10k\Omega$ R7 R4 1.2M Ω VR1 100kΩ subminiature horizontal skeleton Capacitors preset 0.01µF C₆ C1-C3 0.01 µF (3 off) Capacitor 68pF C7 0·1µF 470µF 16V elect. C1 1000pF 10pF C8 C₅ Semiconductors **Integrated Circuit** IC1 Type 741 8-pin d.i.l. D1 1N914 IC2 Type 709 8-pin d.i.l.

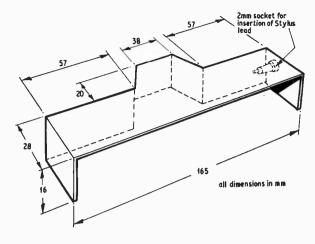


Fig. 3.6. Dimensions of the keyboard cover which was made from 3mm card

The time constant of C7 is such that it will lose only a small proportion of its charge during the negative half cycle of the h.f. signal. The result is that an effectively constant negative signal is presented to the ENVELOPE SHAPERS during the period that the stylus and/or key contacts are made.

ISOLATION RESISTOR

In addition to the components making up the DETECTOR Fig. 3.5 also shows a resistor, R7, in series with the stylus lead and HOLD circuit. The purpose of this resistor is to provide a degree of isolation for C2 in the HOLD circuit so that its relatively large capacity will not over-attenuate the signal on the stylus.

R7 (20kΩ in the prototype) also provides a delay in the d.c. charging rate of C2 with the result that there is a 20ms portamento effect. This effect is not really too noticeable unless consecutive KBD voltages are programmed from opposite ends of the KBD but it could perhaps be a source of irritation for the constructor wishing to use the Minisonic for serious musical purposes.

In these circumstances R7 could be replaced by an inductance which would provide the degree of a.c. isolation required whilst presenting only a nominal resistance to d.c. A suitable choke could be made up from a small ferrite ring toroidally wound with about 20 to 30 turns of 34 s.w.g. enamelled copper wire.

Some experimenting will possibly be required to get just the right value and it would be best to start with the greater number of turns and reduce these as necessary to get the best balance between a.c. isolation and d.c. resistance.

PORTAMENTO

As a modification to the prototype circuits some constructors may wish to incorporate a variable portamento control. In view of the lack of space on the front panel the best way to do this is to mount a miniature edgewise volume control—such as is used on some transistor radios—inside the upper edge of the printed circuit keyboard cover.

The cover is shown in Fig. 3.6. The wire from the stylus socket on the side of the KBD cover would then be routed to one end of the potentiometer while the slider would go via R7, or inductor as mentioned above, to pin 1 on the DIN socket.

RING MODULATOR

The Minisonic RING MODULATOR is an improved version of the circuit which originally appeared in the *P.E. Sound Synthesiser* (August 1973). The essential features of the circuit have been retained however and the circuit is shown in Fig. 3.7.

The RING MODULATOR produces a unique output waveform which comprises, at the same instant, the sum and difference between any two applied input frequencies. This function is carried out in a purpose-built integrated circuit, the SG3402N. With one of the input frequencies fixed, variation in the other will ring the changes in the output frequencies as shown in Table 3.1.

Referring to Fig. 3.7, R1 and R2 form an input attenuator on the so-called carrier input (pin 7) such that, when driven from a vco, the input signal level at C1 will be about 40mV.

Similarly R3 and R4 attenuate the modulator or control input so that, when driven by a vco, the input at C2 is about 200mV. This procedure results in an output signal of about 1.5 volts at pin 4 and the same signal in antiphase at pin 11. The antiphase signals are amplified differentially by IC2 to give a peak output signal of three volts which is then attenuated by R9 and R10 to a level compatible with the remainder of the Minisonic circuits.

SETTING UP THE RING MODULATOR

Setting up the RING MODULATOR is very simple. With the circuit completed link the modulator input to the 0V rail and connect the output to a suitable power amplifier. Apply a signal of about 1kHz to the carrier input (normally connected direct to vco1) and adjust VR1 until the output signal reduces to the lowest possible level. This should, with a correctly wired circuit, be 50dB or more below the peak signal level. At this point the RING MODULATOR is correctly balanced with minimum carrier breakthrough.

NOISE GENERATOR

The NOISE GENERATOR is built round the highly successful ZIJ noise diode manufactured by Semitron Ltd., and is shown in Fig. 3.8. Output from the ZIJ

Table 3.1: OUTPUTS FROM THE RING MODULATOR							
Frequency							
Carrier	700	600	500	400	300	200	100
Modulator	400	400	400	400	400	400	400
Sum	1100	1000	900	800	700	600	500
Difference	300	200	100	0	100	200	300

RING MODULATOR

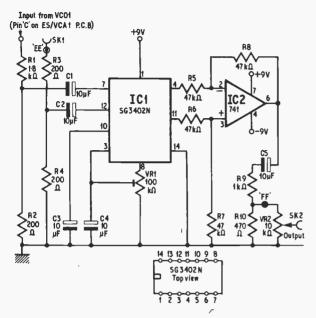
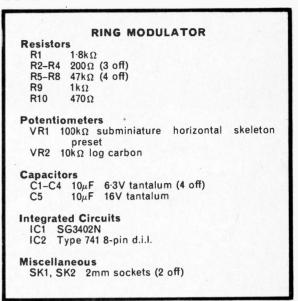


Fig. 3.7. Complete circuit of the RING MODULATOR

COMPONENTS . . .



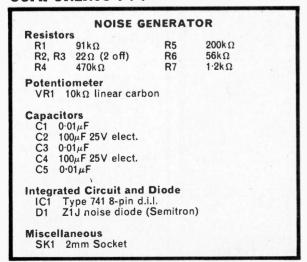
is amplified by the high gain follower IC1 and led, through decoupling capacitor C5, to the volume control VR1.

The NOISE GENERATOR is the only circuit in the Minisonic which does not operate completely successfully down to a battery voltage of \pm 7.5 volts.

In the prototype the noise generator ceased to work when the battery voltage had reduced to $\pm~7.8V.$ This situation may be corrected to a certain extent by shorting out R2 and R3 and/or by reducing the value of R1 to, say, $82k\Omega.$ No setting up is required for this circuit.

NOISE GENERATOR

COMPONENTS . . .



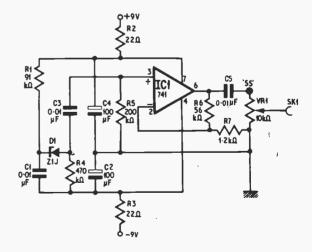


Fig. 3.8. Circuit of the NOISE GENERATOR

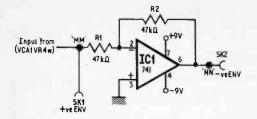


Fig. 3.9. Circuit of the CONTROL ENVELOPE IN-VERTER. This is fed with the output of ES/VCA1 via VR4 (see last month)

COMPONENTS . .

R1, R2

 $47k\Omega$ (2 off) Type 741 8-pin d.i.l. IC1

SK1, SK2 2mm sockets (2 off)

POWER AMPLIFIER

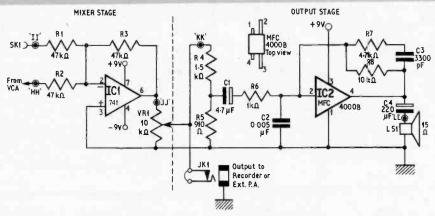


Fig. 3.10. Complete circuit diagram of one of the POWER AMPLIFIERS with integral two-input mixers. Note that the mixer stages are mounted on the main circuit board

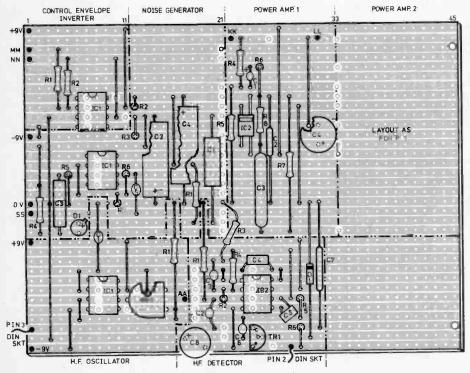


Fig. 3.11. The Veroboard panel which carries the NOISE GENERATOR, HF OSCILLATOR AND DETECTOR, CONTROL ENVELOPE INVERTER, AND POWER AMPLIFIERS

COMPONENTS . . .

POWER AMPLIFIERS AND MIXERS (2 off)

Resistors

R1-R3 47kΩ (3 off)

R4 1.5kΩ

910Ω R5

R6 1kΩ

R7 4.7kΩ **R8** $10k\Omega$

Potentiometer

VR1 10kΩ log carbon

Capacitors

C1 4.7µF 35V tantalum

C2 0.005μF ceramic

C3 3300pF

C4 220µF 40V elect. (or 470µF 16V)

Integrated Circuits

IC1 Type 741 8-pin d.i.l. IC2 MFC4000B

Miscellaneous

LS1 3in 15 speaker

SK1 2mm socket

JK1 3.5mm jack socket

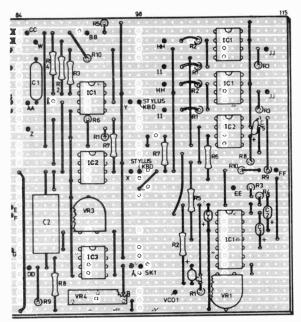


Fig. 3.12. The layout on the main Veroboard panel, the majority of which was shown last month

CONTROL ENVELOPE INVERTER

Shown in Fig. 3.9, the CONTROL ENVELOPE INVERTER represents a modification to the prototype instrument and has been included, principally, so that the VCF may be programmed automatically from ENVELOPE SHAPER 1. The inverter itself is a simple unity-gain inverting amplifier which requires no setting up procedure.

POWER AMPLIFIERS

The complete circuit of the power amplifiers, which includes a two-input inverting mixer, is shown in Fig. 3.10. As with all the virtual earth circuits in the Minisonic the mixer has the minimum number of inputs and almost any number of additional inputs may be applied by following the basic details given in Part 1 of the series.

The slider of the volume control (VRI) at the mixer output is wired directly to a jack socket from which may be taken a signal suitable for driving an external power amplifier, tape recorder, external

mixer, etc.

CIRCUIT BOARD LAYOUT

The CONTROL ENVELOPE INVERTER, HF OSCILLATOR and DETECTOR, NOISE GENERATOR, and POWER AMPLIFIER stages are carried on a separate circuit board which is illustrated in Fig. 3.11.

The KEYBOARD CONTROLLER, RING MODULATOR and POWER AMPLIFIER/MIXER stages are all included on the main circuit board part of which was illustrated last month. The remainder of the board is shown in Fig. 3.12.

Next month: Final wiring-up and adjustments. Keyboard options, as well as circuit additions for more ambitious constructors will be discussed.

Stop Press: The author has developed a printed circuit board to carry all the Minisonic electronics. More details next month.

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ELECTRONICS
FEBRUARY ISSUE ON SALE JANUARY 10, 1975



A selection of readers' suggested circuits. It should be emphasised that these designs have not been proven by us. They will at any rate stimulate further thought. Any idea published will be awarded payment according to its merits. Why not submit YOUR IDEA?

negative going transition at the collector of TR2 is communicated via the bootstrap connection of C2 to the base of TR1. This results in TR1 being cut off and C3 now acts as an effective collector potential for TR2. Therefore C3 discharges through R3 until the emitter potential of TR1 is again negative with respect to the base. Thus TR1 conducts and thereby completes the cycle and the whole sequence will start again.

If the time constants R1-C1 and R2-C2 are sufficiently large, the frequency of oscillation is predominantly determined by C3/R3 and C3/R4. Current consumption from a 9V battery is only 0.7mA since the circuit only consumes current during one half period of the

cycle.

A SERIAL CONNECTION MULTIVIBRATOR

SIMPLE square wave generator with equally fast rise and fall times is always useful. The conventional astable multivibrator has the disadvantage that although one edge of the output waveform is fast, the other edge is comparatively slow. This is due to the fact that the collector of the off transistor has to recover to the potential of the supply in a time determined by the capacitor associated with the collector and the value of the collector load. Improving the output waveform by the use of extra diodes or perhaps an extra transistor are solutions but a novel approach is a serial connection.

Consider the circuit shown and let TR1 be fully conducting and TR2 cut off with C3 fully discharged. It will be evident that C3 has a charging path via R4 and

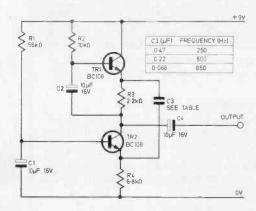


Fig. 1

so this capacitor charges exponentially allowing the emitter of TR2 to approach ground.

When the charging current has decreased sufficiently so that the base potential of TR2 is again more positive than its emitter, then this transistor promptly conducts. This

Unlike its more conventional counterpart the serial connection is always self starting. The output amplitude with the circuit shown is 5.5V peak-to-peak.

M. Harding, Cheadle, Cheshire,

SOUND/LIGHT MODULATOR

A very simple sound/light modulator is shown in Fig. 1 which may be of use to experimenters. Input signals can be taken from the output (loudspeaker) of an audio amplifier since in most pop or disco environments the small amount of distortion introduced using this method will hardly be noticed.

Sensitivity is controlled by VRI whilst the transformer provides isolation and the drive for triac

MAC 11-6. This is possible as the triac drive pulses need not be shaped.

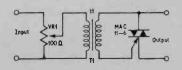


Fig. 1

To provide frequency sensitivity and/or other channels, capacity can be inserted in the triac gate circuit. Using 100V $1\mu\mathrm{F}$ capacitors, bass response can be selected by inserting the capacitor in parallel with the secondary of the transformer. For the treble the capacitor is in series with the gate and for the mid ranges two capacitors are used, one in each of the foregoing positions.

In this way three circuits can be built up to control three separate

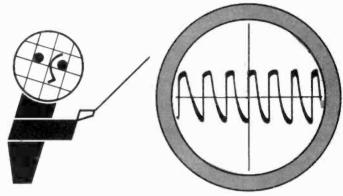
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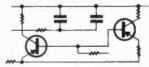
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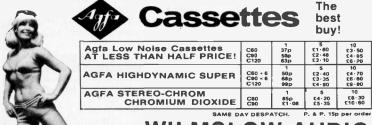
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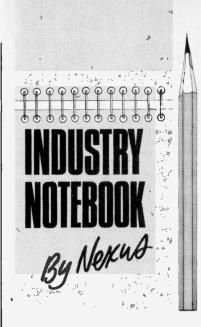
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EASIER SERVICING

Colour television is easily the most complicated box of tricks in the home. Healthy competition has kept prices low, despite inflation. And that same competition, in effect the struggle for market share, is still spurring manufacturers to give the buying public more and more value for money.

It follows that the more complex a machine is the more it costs to service. And TV sets, by their very nature, have a fine in-built monitor, the screen itself, which instantly shows up defects or deterioration in performance. The broadcast authorities are even kind enough to transmit test patterns designed to highlight imperfections.

Every manufacturer has nightmares over servicing. Apart from losing goodwill, every time a set goes wrong during guarantee another chunk of profit goes down the drain. So in their own interests, manufacturers do genuinely try to make sets that are reliable and there has been a strong move over the past few years to build sets in modules to allow servicing by module replacement.

Now Grundig in Germany has gone one step further by building in a rapid diagnosis system in every set. Grundig had switched to a nearly fully modular system some three years ago in which 75 per cent of the circuitry is split among a dozen plug-in modules, each with a defined function. If any one module failed it was fairly obvious to the service engineer and he just plugged in a replacement. Now the diagnostic adaptor enables an instant check of the non-modular circuits.

Key check-points of the circuits are all brought to a single 13-way

socket into which the service engineer plugs his monitor. The monitor has 13 led's and if any one of these fails to light up there is an indication of a specific fault. It costs very little extra on each set to provide the facility and the saving in engineer's time can be enormous. And, of course, with soaring labour costs, time saved is very important, not to mention customer satisfaction. The plug-in diagnostic aid costs the dealer under £10. Quite a bargain. And the customer benefits, too. Other set makers are expected to follow the trend.

EXPORTS

If you've ever thought that export promotion is not given enough priority in Britain, reflect for a moment on the current trading quarter which ends on December 31, 1974. The British Overseas Trade Board is giving support to 1,360 British companies at 76 overseas trade fairs in 27 countries. In addition there are 54 outward trade missions representing 730 companies and involving 45 countries. These group activities are in addition to hundreds of "private" promotions by individual companies.

One of next year's big trade drives is to be centred on Western Canada and seminars are already being held in Britain which will brief exporters on trade opportunities in advance of parties of exporters visiting the two big growth areas of Alberta and British Columbia. In the direct field of electronics, one of our biggest 1974 efforts was at Munich's Electronica Exhibition at which 57 British companies took part in a joint venture.

PACEMAKER BOOM

It was only a few years ago that we were all marvelling at the way microelectronics had made possible the heart pacemaker which has done so much to extend human life. With improved techniques in implant surgery and technical advances in pacemakers this single branch of medical electronics has now blossomed into an industry in its own right with world sales this year expected to top £25 million and reach over £200 million by 1980.

The technical problem which has been engaging pacemaker researchers is how long they can be kept working without recourse to further surgery to replace the battery. One approach was to use nuclear power to give infinite life.

Another was a rechargeable unit that could be recharged by induction through an external unit. But both these solutions are losing favour, mainly expense in the nuclear field and susceptibility to outside interference with rechargeable units, apart from the occasional inconvenience to the user.

It seems now that long-life batteries will do the job quite well. Reasons are that in the early days pacemakers were more powerful than they needed to be, the new active devices using technologies such as CMOS take far less current, and battery technology itself has improved.

With current drain reduced to less than 20 micro-amps a three-year life can be obtained from mercury-oxide-zinc units and possible developments in sodium-bromine and lithium cells could give a battery life of seven years or more. This figure ties in well with the life-expectancy of pacemaker users who statistically have an average age at implant of 67 years and can expect another 5-7 years of life.

HUMBLE HARDWARE

Racks, panels, instrument cases hardly ever hit the headlines. Yet they are still big business in electronics clocking-up European sales of over £30 million a year and double that if you add in PCB edge connectors and other interconnection devices. So don't despise the metalwork in electronics. It might not make so much profit as glamour products but the commercial risk is much lower. Provided, of course, that you can get your materials. One prominent manufacturer, lamenting recently on production hold-ups, commented "screws have seemingly that ceased to exist, and costs have gone up alarmingly". And steel, plastics and paint have also been hard to get.

TAKE OFF

Great sighs of relief that the European Multi Role Combat Aircraft has received the go-ahead for its final development phase. This is the project that is exercising the best brains in electronics. It's costing the earth, of course, but how else do you keep ahead in technology?

At a more mundane level it's good to see Plessey has started delivering ILS systems for Chinese airfields in a contract worth 2850,000. And British aerospace companies as a whole are doing well with exports worth more than £2 million every working day with our best customer still being the United States followed by France and Germany.

MARKET PLACE

Items mentioned in this feature are usually available from electronic equipment and component retailers advertising in this magazine. However, where a full address is given, enquiries and orders should then be made direct to the firm concerned. All quoted prices are those at the time of going to press.

MOUNTING PILLARS

A useful new product for the constructor has just been released by West Hyde Developments Ltd. Called llex pillars, they are designed to insulate circuit boards from cases or chassis and at the same time support them either vertically or horizontally one on top of another.

Made from moulded nylon, they have a rigid girder-shaped supporting section with a spring loop fastener at the top and tension feet together with a push-in clip at the base.

Suitable holes are drilled in the chassis and boards and the pillars are simply pressed into place and any subsequent boards mounted on the top of the pillars, see photograph. The sizes of the pillars vary from \$\frac{1}{2}\$ in and cost approximately \$3p\$ each for a minimum order of 10 (for \$\frac{1}{2}\$ in size).

Full particulars and sizes together with price list can be obtained from West Hyde Developments Ltd., Ryefield Crescent, Northwood Hills, Middlesex, HA6 1NN.

ON TAPE

Two new cassette tapes have been announced recently by EMI Ltd and 3M United Kingdom Ltd.

The new X1000 ferric oxide cassette from EMI is claimed to give as good reproduction as chrome dioxide cassettes. A C60 cassette is expected to retail at 99p (excluding VAT).

The main technical improvements claimed for the X1000 are: an increase of 3-4dB in the 8-15kHz frequency range, compared to low noise tapes. A wider dynamic range due to the tape's increased magnetic remanence, resulting in less tape hiss. Improved high frequency response ensures a low level of intermodulation distortion.

When used with good quality audio equipment the tape is claimed to give excellent performance down to 25Hz and up to 15kHz.

Undoubtedly the new Sco

Classic tapes from 3M's, with a C90 cassette at £2·16, is aimed at the "serious" end of the market.

This new double coated or dual layer tape indicates the trend

towards the use of a product compatible with existing tape and equipment rather than the current trend of using metallic dioxide tape. It is claimed that the new tape combines the high frequency abilities of chromium dioxide with the bias characteristics and low frequency response of the low-noise ferric oxide tape.

Both the above tapes are available from all good audio shops and large stores.

LITERATURE

A comprehensive 724-page data book covering Motorola's range of linear i.c.s is available from Semicomps Ltd. The book contains not only full data but, in many cases, valuable application information on over 300 devices.

The range includes op. amps, drivers and line receivers, d/a and a/d converters, comparators, voltage regulators, timing and power control units, consumer TV, audio and radio circuits, r.f. amplifiers and automotive circuits.

For easy reference the data sheets are arranged in alpha numeric sequence without regard to product category.

The book costs £1.26 and is obtainable from Semicomps Ltd., Northfield Industrial Estate, Beresford Avenue, Wembley, Middlesex, HAO 1SD.

NEWS BRIEFS

Readers who are building the "P.E. CCTV Camera" may be interested to know that Crofton Electronics are now able to offer a complete kit of parts for this project. They can also supply lenses,

coils, tubes and printed circuit boards separately.

For full details readers should write to Crofton Electronics at 124 Colne Road, Twickenham, TW2

GQS.

We understand that Re An Products Ltd are able to supply all the control knobs (19 with skirts and one without) for the P.E. Minisonic synthesiser. These knobs have a translucent numbered skirt and are available with coloured caps.

It has been suggested that by using a colour code system for the knobs the front panel layout of the Minisonic can be identified in colour groups (i.e. envelope shapers, voltage controlled oscillators and amplifiers, etc.), which can make the instrument easier to use.

A price list for the knobs, type R62, is obtainable from Re An Products Ltd, Burnham Road, Dartford, Kent, DA1 5BN.

What is believed to be a unique service for the private constructor, has been announced by SCS Components.

Now, branded guaranteed components are being offered at very competitive "one-off" prices, in fact the same as applying to industrial users. Included in this offer is a very large range of integrated circuits and transistors.

A complete price list (free) is obtainable from SCS Components, Northfield Industrial Estate, Beresford Avenue, Wembley, Middlesex.

To help beat rising costs Amtron U.K. Ltd. are now able to supply direct to the customer many of their more expensive electronic kits.

A full list of the construction kits available is obtainable from Amtron U.K. Ltd., 4 Castle Street, Hastings, Sussex TN34 3DY.



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KIT 18: 2 way and crossover (1 \times 8" plus 1\frac{1}{2}" dome) (15-20 litre box) 30 watts 35-20,000 c/s. £16.30.

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Practical Electronics January 1975



ROBOTICS

By John F. Young Published by Butterworths 300 pages, $8\frac{7}{2}$ in $\times 5\frac{1}{2}$ in. Price £6.00

THERE is nothing mythical about this book. It is concerned with hard facts of engineering. Robots have a very real existence, and in various forms perform many useful tasks in industry and elsewhere. Tasks which range from the mundane and repetitive to those

of a highly specialised and skilled nature.

All this is clearly brought out in Robotics. The author, who has had a long experience in this field, describes notable developments in robot design which have been or are currently being undertaken in various advanced countries. Some of the devices mentioned will be familiar, like the Lunakhod, the Robotug, and the mobile robot used by the army to explode bombs left in cars. But there are many other significant developments in robotics which this book now brings to the attention of a wider audience.

Underlying all this activity are unmistakable signs of the eventual appearance of the General Purpose Robot for domestic use. The idea of an automated housewife is not just wishful thinking; it is the target of many designers in a number of countries. Indeed, according to the author, the Japanese are already well advanced with plans for a G/P Robot suitable for mass production with the inevitable economic advantages this will bring. This book does not cover the "brain" of the robot,

but concentrates upon the engineering of the "body' Hydraulic, pneumatic, and electrical techniques for actuation of the "limbs" are discussed. Electronics comes into its own with the imitation of the human senses, and all the commonplace sensoring devices are considered. Character and voice recognition pose greater problems; some indication of recent work by the author and his associates at Aston University in devising circuitry to solve some of the difficult problems in this frontier area of electronics is given.

Extensive lists of references accompany each chapter. Valuable as sources for the specialist researcher, these items provide additional evidence of the determined efforts that have already been made in this field of Robotics, and indicate how the robot has already become usefully employed in so many and varied everyday

activities.

ELECTRONICS - AN ELEMENTARY INTRODUC-TION FOR BEGINNERS (SI UNITS)

Bv L. W. Owers Published by Publication Mailing Services 120 pages. Price £1.45

N LINE with the current tendency, this book is perfect bound between linen covers to keep costs to a reasonable level. It is nonetheless a valuable introduction to electronics for any beginner, particularly those meeting this type of subject for the first time.

Diagrams are used extensively with the addition of formulae where necessary and these latter are spelt out

in SI units to conform to international practice.

The reader is led by the hand through complexities of fundamental particles, atoms, energy in its various forms, static and current electricity and the basic raw materials of electronics from theory to simple example.

Finally, the theory is exemplified by discussions of the valve and semiconductor and their use in radio and

television and other areas.

NEWS BRIEFS

G. D. SHAW LECTURES AT AUDIO FAIR

HE tremendous interest in synthesisers was reflected by the massive attendance at the two lectures at this year's Audio Fair by G. D. Shaw, the author of P.E. Minisonic articles currently appearing in *Practical* Electronics.

Entitled "Sound Synthesis for the Amateur", the lectures described synthesisers ranging from the simplest. in the form of the Minisonic, to the synthesiser of the future in the form of a digitally organised instrument having full polyphony (the ability to play more than one note simultaneously), and a memory facility.

The part of the lecture dealing with the Minisonic was illustrated with some impressive tape recordings made, using the Minisonic, by Malcolm Pointon. Most people were amazed at the range of effects that could

be produced by such a simple instrument.

In the realm of digital synthesisers, Mr Shaw hopes to be the first to produce a design suitable for the amateur, a formidable task when one appreciates the complexity of such a system. The instrument is to be designed in such a way as to allow expansion from a basic unit simply by plugging in printed circuit boards as and when they are needed (or can be afforded).

This was not the only area where P.E. scored a "sound" success. On our stand the "P.E. Joanna" piano, exhibited for the first time, created enormous interest amongst the public. This unique instrument, ideal for the modern home, features piano, harpsichord and honky tonk facilities with true touch sensitive operation.

The P.E. Joanna is a future project and full details

will be published in the next few months.

LINK(S) UP

N almost unbelievable 5 watts of transmitted power A was used, in conjunction with a satellite, to beam a transmission over a distance of more than 50,000

miles recently in America.

In an experiment involving a simple antenna made from a golfer's umbrella, an engineer from the General Electric Company of USA used a low power "walkietalkie" radio to prove that even with such rudimentary equipment the only important requirement for long-range communication is the presence of a satellite overhead.

To be fair, the antenna was specially made up from the golfer's umbrella but nonetheless a morse message was beamed from the NASA headquarters to a geostationary satellite ATS-3 and then to GE (USA)'s Radio-Optical Observatory near Schenectady, New

York.

The demonstration shows all too clearly just how easy it would be for almost world-wide coverage to be provided for some form of search-and-rescue system based on a simple man-carried emergency transmitter. It is also envisaged that the system could carry phone signals, not just morse.

PRICE CUT

N this period of constantly rising prices it is heartening to know that at least some items are becoming cheaper. Motorola have recently announced their second

price reduction in the CMOS device area.

This second reduction, worth an average of 25 per cent, applies to standard MC14000 and the in-house MC14500 devices. When applied to MSI the new pricing will give an individual gate function cost at around just a few pence and when this is coupled to the saving in power supply requirements and package count the total effect is a distinct improvement when using CMOS.

REGULOTTE— A SELECTION FROM OUR POSTBAG

Gas Detectors

Sir—As the comment appearing on page 794, September 1974 issue, may raise doubts as to the life of the TGS sensors, we trust the following brief outline of the operating principles of the sensor will indicate why we claim that their life is comparable to that of other semiconductor devices, rather than the catalytic type of gas detector.

Molecules of flammable or de-

Molecules of flammable or deoxidizing gases are absorbed on the surface of the Taguchi sensors, resulting in electron transfer between absorbate and the solid sensor surface. In the case of hydrocarbon gases the reaction is related to the ionization potential of the gas absorbed on the surface of the pellet. The lower the ionization potential, the more readily is the gas detected. Hence isobutane (ionization potential 10.79eV) is detected more easily than methane (ionization potential 13.04eV).

The change in conductivity of the sensor is not caused by heat resulting from the combustion of a gas at its surface. The lack of combustion and relatively low operating temperature, 250°C, eliminates deterioration of the inert 82 per cent Palladium + 18 per cent Iridium filaments encapsulated in the bead.

The Taguchi gas detectors have been in continuous use in Japan for six years, and the only noticeable change in performance has been an increase in sensitivity with time, up to a maximum of 30 per cent when a levelling-off occurs.

Damage can occur to the sensor if it is exposed for long periods to high concentrations of gases containing sulphur or lead, and such gases will in any case inhibit the performance of most gas detectors.

There are of course, many applications where the catalytic type sensor is superior to the Taguchi especially in the areas of selectivity and long term repeatability, but where a pre-set low level long-life sensor is required, the Taguchi have found good acceptance. More than two million have been put in service to date.

From the design point of view, it is essential that a current limiting resistor be included in series with the sensor, or else the sensor may

be destroyed by excessive current at switch-on. For example, if a low voltage 6V to 24V circuit is used then the minimum value of load resistor is $2 k\Omega$. A variable resistor alone should not be used in this position as it is possible for the unit to be switched on with the variable resistor set to the low end of its range.

D. Lahiff, Manager, Figaro Engineering, Shannon, Ireland.

Gas Sense (or)

Sir—In the circuit of the "Boat/ Caravan Gas Detector" (October 1974) the fact that the l.e.d. is alight proves that the heater side of the gas detector is connected and conducting. A dangerous situation could arise if the alarm circuitry, or more likely the wire connecting it to the detector, were to become opencircuited. This problem could arise from corrosion of the B7G socket or the gas detector pins (and boat bilges are known to be very damp places).

The simple modification shown in the enclosed diagrams will enable the detector to be fully tested each time it is turned on. The resistor R16 corresponds to a concentration of approximately 0.2 per cent of Butane or Propane in air, well below the inflammability range of 1.8 to 9 per cent.

R. A. Wood, Wolverhampton.

Growing Upwards!

Sir—With reference to Mr. Crilly's letter (*Readout*, September), I feel that I in turn must draw attention to two points.

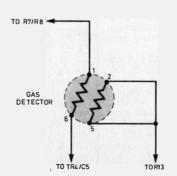
Firstly, the magnetic properties of a material are determined by the spin configurations of the constituent atoms or ions. The vast majority of elements and compounds are said to be paramagnetic—this means that unbalanced spins among the electrons leave a nett magnetic moment on the atom. In the presence of a magnetic field, at low temperatures so that thermal vibrations do not upset things, a degree of alignment can occur.

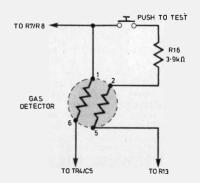
In the case of iron and certain other materials, ferromagnetism is observed. This means that over small regions of the crystal the magnetic moments (which are due to the same mechanism as above) are aligned by an internal crystal field. These regions are called "domains", and may themselves be aligned by an external field. Thus, ferromagnetism is a bulk property of iron, and it is inaccurate to speak of ironcontaining molecules as being more "magnetic" than magnesium containing ones without considering the nett spins for the molecules concerned.

Secondly, Mr. Crilly states that strong magnetic forces radiate from the centre of the earth. This is not strictly true; the earth's field is not a monopole but approximates to a dipole. As a result of this there are regions of the earth's surface at the geomagnetic equator, where the field is parallel to the surface—in fact it is vertical only at the geomagnetic poles.

The suggested mechanism of tropisms would, therefore, not work in general, producing horizontal roots at the geomagnetic equator, and roots inclined to the vertical in most regions. I should also point out that plants transplanted from the Northern hemisphere to the Southern would develop into roots growing upwards at an angle to maintain their accustomed orientation with respect to the field.

C. R. Francis, Sheffield.





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VA30 CHANNEL £3.50 Carr. free
This is the basic channel module in the above mixers and may also be used for extra inputs on either the mono or stereo mixers. Fitted with volume, bass and treble controls, requires just a jack and supply

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Parapsychic pnenomena

Sir-Sometimes I read your articles about "ESP" in PRACTICAL ELEC-TRONICS with great interest. I'm working in parapsychic phenomena as a "hobby" and I have some results, which may interest you and your readers.

During the spring and summer I discovered that a magnetic field stimulates germination and even growth of seeds and plants. I made experiments with lobelia, bindweed, pea, lentil and various other plants. In all cases germination increases about 20-30 per cent, germination time was shortened by approximately 30 per cent and plants were 50 per cent higher than in control tests. In addition, it is interesting that the same effect was caused by hand movement, so called "magnetisation", known 200 years ago and used by Mesmer for healing with "animal magnetism"

It is my opinion that these effects are caused by water polarisation in both cases. The same effect is obtainable by using "magnetised" water, activated by passing through a strong magnetic field.

These two phenomena will be discussed and published in two parapsychological magazines.

V. Patrovsky, Czechoslovakia.

A Boolean Breakfast

Sir-I was recently having breakfast in an hotel and reading a book on Boolean Algebra. I had reached a point where the author stated that $A + (A \cdot B) = A$, when I realised that the waitress was looking over my shoulder. She asked whether I would like egg, or egg and bacon. My natural gluttony lead me to order egg and bacon. I continued reading and learnt that the expression above was read as "A or A and B equals A", and that if either A or B is present it is given a value of I and if absent its value, reasonably enough, is 0. The author proved the statement by a truth table.

At this moment the waitress placed in front of me a plate on which there was an egg but no bacon. Of course, I pointed out with maximum natural charm that I had ordered egg and bacon. I was surprised, even dismayed, when she said that what she had served was the same as egg and bacon, the book had just proved it by a truth table-egg, or egg and bacon equals

Later, when the management presented me with the bill for bed and breakfast, I pointed out that the bill for bed, or bed and breakfast, should be the same. In the tariff, breakfast was a separate item, and although I had just eaten it, like the

bacon I could prove by Boolean Algebra that it did not exist. One could not be charged, even in these times of rampant inflation, for something that did not exist whether one ate it or not. The cost of breakfast should therefore be deleted.

The Manager appeared to have an entirely mistaken grasp of Boolean Algebra, and the adviser he called in had not even heard of it. Possibly Boolean Algebra does not form part of a constable's training.

Perhaps some of your more erudite readers can point out where the fallacy lies.

R. Parfitt, Croydon.

Make an offer

Sir-Before my husband's death in May 1974 he had started to buy various parts for the "Electronic Piano", described in your Magazine some time ago. He already bought two manual contact assemblies, two C-C keyboards, digital master oscillator and other parts. Is it possible for you to help me to dispose of these items?

Mrs. E. Szwimer.



ELECTRONICS LTD 58-60 GROVE ROAD, WINDSOR, BERKS.

FAST SERVICE. IC's

SEND C.W.O. ADD VAT TO ALL PRICES IN U.K. P&P 15P. BUROPE 25P.OVERSEAS 65P.



0-9DP fl.15 ea LED 0.3" digit 0-9DP f1.49 ea JUMBO LED 0.6" 0-9DP £2.25 ea LIQUID CRYSTAL 6 digit £18

LEDS 14P./ MINI PIN SOURCE OR RED DIFFUSE LEDS. 209 STYLE. NO CLIP. 14P ea

TIL209 RED LED & CLIP 17P ea BIG 4" RED LED & CLIP 18P ea ORANGE & GREEN LEDS: MINI 25P ea.BIG & CLIP 33P ea INFRA RED LED 11.2N5777 33P. PS12 PHOTO 1C/amp/switch 11.

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MOS INTEGRATED CIRCUITS. AY51224 4 DIGIT CLOCK supplied with 14pin socket & data 14.25 MM5311/14 6 DIGIT CLOCK with 28 pin socket & data 31DIGIT DVM AY53500 £7.50 4DIGIT COUNTER/DRIVER £7.50

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As used in imported types costing f100.0nly requires a case & electronics.Heads supplied. Send for data 15p

702 OPA 703 RF/IF 69p 28p 709 T099 709 DIL 14 710 DIL 14 .36p 720 Radio £1.39 723 Regulator67p 741 T099 29p 741 DIL8

741 DIL14 31p 747 Dual 741 89p 748 DIL 8 36p 1505 IC A/D Converter £7 7805 1A5V £1.59 7808 1A8V £1.69 7812 1A12V £1.69 7812 1A12V 11.69 7815 1A15V 11.69 76009 1W AF 75p 76013 6W AFI1.39 8038 Sig Gen 13 69p CA3046 CA3046 69p LM301 OPA 49p LM307 OPA 49p LM308 HiBoPa 95p LM309K Reg.£2.29 LM371 RF/IF £2

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1A50Vrect 4p ea IN914 4p Price each: AC127/128 16p AC187/188 19p AD161/162 35p BC107/8/9 9p BC132/4/7 18p BC147/8/9 10p BC157/8/9 12p TIP2955 TIP3055 TIS43 UJT 25 D IN4001 1N4004 IN4148/914 2N697 13p 10p BC167/8/9 12p BC177/8/9 18p BC182/3/4*11p BC212/3/4*12p 2N706/8 2N2926royg *A or L BCY70/1/2 15p BD131/2 39p 2N3O53 2N3055 BD131/2 35p BFY50/1/2 18p 17p 2N3614 2N3702/3 2N37O4/5 2N37O6/7 2N37O8/9 MJE2955 95 p 62 p M.TE3055 2N3710/11 49p 8p MPU131put 2N3563/64 16p 2N3566/67 16p 2N3638 16p 2N3641/2 16p 0A91

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2N3819E

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CAPACITORS 22pf to O.luf 4p ea.ELECTROLYTIC 25V 2/10/50/100uf 6p.1000uf 20p PRESETS VERT:5p.RESISTORS5% 12p

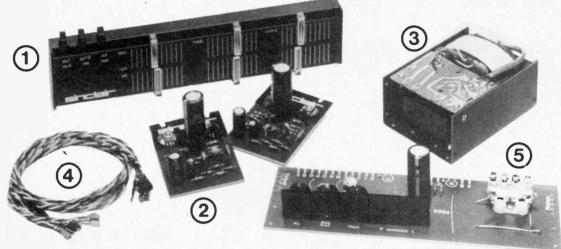
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PROFESSIONAL GOLD PLATED & GREY NYLON. 8,14 or 16 PIN ONLY 15p each.

Forward with Project 80 into



Everything you want in one pack to build the world's most advanced modular hi-fi WITHOUT SOLDERING

- 1 Stereo 80 Control Unit For mag. and ceramic cartridges, radio and tape.
- 2 Project 80 power amplifiers Two Z.40s to give 8/8 watts R.M.S output per channel.
- 3 Power supply unit One PZ.5.
- 4 Connecting wires
 All wires plus nuts, bolts, screws etc.
- 5 Project 805 Masterlink For input and output connections.
- 6 Mains switch block and instructions manual (not illustrated).



SINCLAIR RADIONICS LTD London Rd, St. Ives, Huntingdon PE17 4HJ Telephone St. Ives (0480) 64646

This is Project 80 made even easier to build

You have seen how the marvellously compact Project 80 modules (only 2" high $\times \frac{3}{4}$ " deep) are so adaptable and easy to install. Now, with Project 805, this wonderful system is made easier still to put together. In this, you have not only all the Project 80 modules in one pack for building an 8/8 watt R.M.S. hi-fi amplifier — there is also a loom of colour coded wires cut to length and tagged for clipping on so that you don't even have to solder! Input and output connections go via the 805 Masterlink panel. With the explicit stage-by-stage large 32 page instructions manual included, it becomes easy for anyone, no matter how inexperienced to install an ultra-modern assembly so advanced in appearance and design that it sets brand new concepts in domestic hi-fi — and of course, you can convert to quadraphony just whenever you wish by adding 805SQ. Only Sinclair know-how and manufacturing facilities could hope to bring you such quality and versatility.

TAGGED WIRES CUT TO LENGTH NO SOLDERING

Project 805

the complete ready-to-build hi-fi STEREO AMPLIFIER

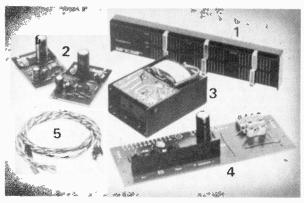
Project 805 comprises a Stereo 80 Pre-amp/Control Unit with input for both magnetic and ceramic cartridges, radio, tape; separate bass and treble cut/lift, and volume controls 2 × Z.40 power amplifiers, PZ.5 power unit, 805 Masterlink, wire loom, instructions manual, etc. down to nuts, bolts and washers. For technical specifications, see third page of this advertisement.

£39.95

+£3.20 VAT (R.R.P.)



true quadraphonics... NOW!



The most effective and economical way to enjoy this spectacular break-through in hi-fi listening

- 1. Project 80SQ decoder with controls.
- 2. Two Z.40 power amplifiers.
- 3. PZ.5 power pack.
- 4. Project 80Q Masterlink unit.
- 5 Wire loom, with clip-on tags NO SOLDERING!
- 6. (Not illustrated) Instructions manual, nuts-bolts, washers, etc.

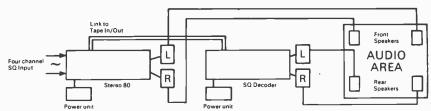
Add a fourth dimension to your stereo sound

It's so simple to convert to quadraphonics when you already have Project 80, or are about to start with Project 805. Project 805SQ is a complete add-on system at the heart of which is the Project 80SQ decoder. It uses the CBS.SQ matrix principle, by now the widest used method of containing four sound channels within the groove of the record. Project 805SQ includes two power amplifiers, power supply unit, connecting wire loom, 805Q Master-link, switch block and instructions manual. The 80SQ decoder (also obtainable separately) has independent tone and volume slider controls on the two rear channels for matching true four channel sound to domestic environment. Project 805SQ is money saving too since you do not have to scrap existing Project 80 equipment to enjoy the newest and most exciting form of home listening in the entire history of sound, and your Project 80 quadraphonic assembly is compatible with stereo and mono records.

- Frequency response ± 3db 15 Hz-25kHz
- Rated output 100mV
- S/N ratio 58dB
- Distortion 0.1%
- Power requirements 22-35 volts
- Phase shift network 90° ±10,100 Hz-10kHz
- Adaptable to discrete (CD4) use



Project 805SQ



The output from any good stereo cartridge feeds into Stereo 80 and passes via the tape outlet to the 80SQ decoder. Here the signal is separated into its constituent 4 channels, those for the front being accepted by the Stereo 80, those for the rear going from the decoder to the two additional power amplifiers and speakers.

£44.95

+£3.60 VAT (R.R.P.)

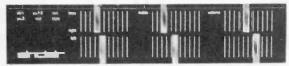
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Project 80 quadraphonic modules may be purchased separately if required. The Project 80SQ decoder may be used with any other amplifier having tape and monitoring facilities. Z40 or Z60 power amps can be used as required.

The Project 80 programme to date

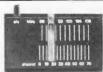
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Stereo 80 pre-amp/control unit



260 x 50 x 20mm (10½ x 2 x ⅔ ins.) separate slider controls on each corrected) ceramic – 300mV, Radio 100mV, Tape 30mV S/N ratio 60dB. Frequency range - 20Hz to 15KHz ± 1dB. OUTPUTS - 2.5V rms max (30V. supply) and tape plus AB monitoring. PRESS BUTTONS for P.U., Radio and Tape. Operating power - 20 to 35V. Black case with white £13.95 +£1 12 VAT (R.R.P.) indications

Project 80 F.M. tuner



Size $85 \times 50 \times 20$ mm ($3\frac{1}{2} \times 2 \times \frac{3}{4}$ ins.). Tunes $87 \cdot 5$ to 108MHz. DETECTOR – I.C balanced coincidence (I.C equivalent to 26 transistors) Distortion - 0.2% at 1KHz for 30% modulation. SENSITIVITY - 5 microvolts for 30dB quieting. Output - 300mV for 30% modulation. Aerial imp. - 75 Ω or 240-300 Ω . Dual Varicap tuning. 4 pole ceramic filter. Switchable A.F.C. Operating power 23-30 volts.

£13.95 +£1.12 VAT (R.R.P.)

Project 80 stereo decoder

Size 47 x 50 x 20mm. For adding to Project 80 FM tuner. With one I.C equal to 19 transistors, and LED indicator which glows on tuning in stereo signal.

£8.95 +72p VAT (R.R.P.)



Project 80 active filter unit (A.F.U.)



Size 108 x 50 x 20mm. Useful where there is need to eliminate unwanted high frequencies (scratch, whistle, etc) or low (rumble). Voltage gain minus 0-2dB. Frequency response (filter at zero) 36Hz to 22KHz. H.F. cut (scratch) variable from 22KHz to 5.5KHz 12dB/octave slope. L.F. cut (rumble) - 28dB at 28Hz, slope 9dB/octave.

£7.45 +60p VAT (R.R.P.)

Project 80 power amplifiers

Intended for use in Project 80 installations, these modules readily adapt to an even wider range of applications. Both incorporate built-in protection against short circuiting and risk of damage from mis-use is greatly reduced

Z.40

80

Size - 55 × 80 × 20mm

9 transistors

Input sensitivity - 100mV

Output - 12 watts RMS continuous into 8 Ω (35v)

Frequency response - 10Hz - 100KHz ± 1dB

S/N ratio - 64dB

S/N ratio = 64dB Distortion = 0.1% at 10 watts into 8 $\,\Omega$ at 1 KHz $\,$ £5.95 $\,^{+48p}_{VAT\,(R.R.P.)}$



To Sinclair Radionics Ltd. Please send, (carriage paid in U.K.)

Size - 55 × 98 × 20mm

12 transistors

Input sensitivity - 100-250mV

Output - 25 watts RMS continuous into 8 \O(50V)

Distortion - 0.02% at 10W/8 Ω/1KHz

Frequency response - 10Hz to more than 200KHz ± 3dB

S/N ratio - better than 70dB

Built-in protection against transient overload and short circuiting Load impedance - 4 \(\Omega\) min; max, safe on open circuit

£7.45 +60p VAT (R.R.P.)

Power-supply units

PZ.5 Unstabilized, 30 volts. Suitable for Z.40 assemblies, etc.

£5.95 +48p VAT (R.R.P.)

Stabilized. Output voltage adjustable between 20 and 50 volts approx. Protecting fuse.

£8.95 +72p

PZ.8 Stabilized. Output adjustable from 20 to 60V. approx. Reentrant current limiting makes damage from overload or even shorting, impossible. Without mains transformer.

f8.45 +68p VAT (R.R.P.)



Project 805 (previous pages)

£39.95 +£3.20 VAT (B.R.P.)

Project 805SQ quadraphonic

add-on kit £44.95 +£3.60 (B.B.P.)

Project 80SQ quadraphonic decoder



Size 260 × 50 × 20mm, matching Stereo 80 in style. Connects with tape socket on stereo 80 or similar facility on any stereo amplifier. Frequency response 15Hz to 25KHz + 3dB. Distortion 0.1%. S/N ratio 58dB, Rated Output - 100mV. Separate bass and treble slider controls on each channel, also volume. Phase shift network 90° ± 10, 100Hz to 10KHz. Operating power - 22-35V.

£18.95 +£1.52 VAT (R.R.P.)

Sinclair Q.16 loudspeaker

An original and uniquely designed speaker of outstanding efficiency. Balanced sealed sound chamber and special driver assembly. Loads up to 14 W./R.M.S. 8 ohms imp. Size 248mm square × 120mm deep. Pedestal base. All-over black front, teak surround.

£8.95 +72p VAT (R.R.P.)



Sinclair Radionics Ltd., London Road St. Ives Huntingdonshire PE17 4HJ Telephone St. Ives (0480) 64311

R.O. St. Ives: Reg No. 6994583 Eng.

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SUPPLIERS OF QUALITY PRINTED CIRCUIT BOARDS, KITS AND COMPONENTS TO A WORLD-WIDE MARKET

The ever-popular AURORA—4 or 8 channels each responding to a different sound frequency and €controlling its own light. Can be used with most audio systems and lamp intensities. A must for any Disco, and a fascinating visual diaplay for the home.

4 channel component set (excl. thyristors)	£11-49
8 channel component set (excl. thyristors)	£20 · 32
Power supply component set	£4 - 78
PCB for 4 frequency channels	£2·50
PCB for power supply and 8 lamp drivers	£1·25
PCB for power supply and 8 lamp drivers	£1·25

P.E. CCTV CAMERA Details in List

VOICE OPERATED FADER

For automatically reducing music volume during "talk-over"—particularly useful for Disco work, or for home-movie shows.

Component set, incl. PCB £2.95

P.E. GEMINI 30W STEREO AMPLIFIER

An exceptionally high quality Stereo Amplifier system, specifications for which are shown in detail in our list, together with semiconductor requirements.

Main Amplifier: Set of resistors, capacitors and presets Stereo printed circuit board	£5 · 96 £1 · 28
Pre-Amplifier:	
Sets of resistors, capacitors, potentiometers	
and switches-	
Standard Tolerance Set	£10 · 57
Superior Tolerance Set	£16 · 04
	£2 · 20
Stereo PCB (as Published)	12 - 20
Requisted Power Supply:	
Set of resistors, capacitors and preset	£4 - 58
Deleted elected based	72p
Printed circuit board	/4p

HI-FI TAPE LINK

Designed for use with reasonable quality tape decks, this high performance pre-amp includes record, playback and metering circuits.

Stereo component set (excl. panel meter) Mono component set (excl. panel meter)	£22 · 05 £13 · 31
Power supply component set	£3-72
Stereo main PCB	£2 · 50
Stereo sub-assembly PCB	86p

TAPE-NOISE LIMITER

Very effective circuit for reducing the hiss found in most tape recordings.

Component set (incl. PCB)	£2 · 30
Regulated power supply (including printed circuit board)	£3·71

PROJECT Q4

Multi-system Quadraphonic Decoder

Decoder component set	£13·74
Power supply components	£3·22
Set of PCBs	€2 - 60

SEMICONDUCTOR TESTER

Essential test equipment for the enterprising home constructor.

Set of resistors, capacitors, semiconductors, potentiometers, makaswitches and sub-assembly PCB (fuller details in list)

A simple but effective manually controlled unit for introducing the "phasing" sound into live or recorded

Component set (incl. PCB)

P.E. SOUND SYNTHESISER

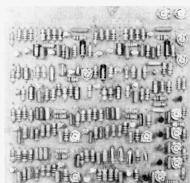
The well-acclaimed and highly versatile Synthesiser published in P.E. Feb. 1973 to Feb. 1974.

Component sets and printed circuit boards: Full details

RHYTHM GENERATOR

Programmable for 64.000 rhythm patterns from 8 effects circuits (high and low bongos, bass and snare drums, long and short brushes, blocks and cymbal), and with

Tempo, Timing and Logic Circuit Component set (excl. switches) Double-sided PCB for above	£17 · 25 £2 · 30
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Monitor Amplifier Component set and PCB Power Supply	£3·10
Component set and PCB	£5 ⋅ 65



AUDIO MILLIVOLTMETER

Wide-ranges and good accuracy. Component set (excl. meter) while stocks last

ULTRASONIC TRANSMITTER-RECEIVER

A highly sensitive and long range "invisible detection circuit with numerous applications. Component set with PCBs, but excluding trans-

P.E. RONDO

Minitron 3015F

225p

PCB details in List.

P.E. ELECTRONIC PIANO

Details in List.

HOME INTERCOM Details in List.

POWER SLAVES

€5 - 23

SOUND RENDER

A multi-purpose sound controller, the functions of which include envelope shaper, tremolo, voice operated fader, automatic fader and frequency doubler.

Component set Printed circuit board

REVERBERATION UNIT

A high-quality unit having microphone and line input pre-amps, and providing full control over reverberation level.

Component set (excl. spring unit) Printed circuit board

P.E. MINISONIC Details in List

8W AMPLIFIER

A moderately powered amplifier of more than average

Main Amplifier	
Mono component set	£4 - 18
Stereo component set	£8 · 36
Mono printed circuit board	72p
Pre-Amplifier	
Mono component set	£2·50
Stereo component set	26 - 46
Stereo PCB	21 - 65
Power Supply	
Component set	08-63

BIOLOGICAL AMPLIFIER

Multi-function circuits that, with the use of other external equipment, can serve as lie detector, alphaphone, cardiophone, etc.

Pre-Amplifier Module Component set and PCB	£3 · 48
Basic Output Circuits Combined component set with PCBs, for alpha-	
phone, cardiophone, frequency meter and visual feed-back lamp driver circuits Audio Amplifier Module	£4·96
Type PC7	£5·20

PHOTOPRINT PROCESS CONTROL

For colour and B & W, an indespensible dark-room unit for finding exposure, controlling enlarger timing, and stabilising mains voltage.

Component set (excl. meter) Printed circuit board	£8 · 85 £1 · 60

ENLARGER EXPOSURE METER AND THERMOMETER

Dual-purpose dark-room unit with good accuracy.

Component set with PCB, but excluding meter

WIND AND RAIN UNIT

Component set incl. PCB

A manually controlled unit for producing the above-name

OVERSEAS COLOUR CODE PCB LAYOUT AND P. & P. will be charged at cost. VAT does not currently apply. List Add 8% (or current rate if different) to Send S.A.E. for free list CIRCUIT DIAGRAMS SUPPLIED WITH ALL giving fuller details of kits, PCBs, and other SUPPLIED WITH MOST KITS AND AS PART OF gives fuller details including kit Add 18p to all orders PCBs DESIGNED BY order cost inweights. Charge for list: Europe 10p, other countries 20p. cluding P. & P. components. PHONOSONICS Semiconductors Integrated Circuits Polyester Tantalum

AC128 AC176 BC107 BC108 BC109 BC147 BC148 BC149 BC157 BC158 BC159 BC182L BC204 BC204 BC204 BC212L BC478 BC212L BC478	28p MJE3055 28p NKT0033 13p OC28 13p OC28 13p OC84 12p CTX107 12p ZTX107 12p ZTX503 13p ZYX513 13p ZYX513 13p ZYX51 13p ZYX51 14p ZYX51	75p 2N3822E 112p 2N4870 65p 2N4871 14p 2N577 14p 2N577 15p 1N914 15p 1N4001 13p 1N4002 13p 1N4002 13p 1N4002 22p 1OD6 27p BA145 27p OA91 22p 0A90 36p 1GP7 50p 1SJ50	36p 7; 38p 7; 45p 7; 4p 7; 6p 7; 7p 7; 8p 7; 8p 7; 23p 1; 23p 1; 12p C	23 T O5 41 8-pin DiL 41 8-pin DiL 41 14-pin DiL 42 T O5 48 8-pin DiL 48 14-pin DiL 400 402 420 420 447 473 1 7473 2 1,78715 T O3 2 2,78715 T O3 2 2,30464 2 2 3,30464 2	15p 4.7V 63p 5.6V 63p 6.2V 63p 9.1V 20p 11V 20p 12V 75p 18V 44p 18V 44p 20V 550p 20V 550p 27V 68p	400mW 12p 1W 25p 1·3W 20p 400mW 15p W 20p 00mW 15p W 20p 00mW 15p 00mW 15p 00mW 15p 00mW 15p 00mW 15p	0 · 47/63V 1 0/63V 1 · 5/63V 2 · 2/63V 4 7/63 6 · 8/40 10/25 10/63 15/40 22/10 22/25 33/6 · 3 33/16 33/50	Capacitors (uF/V) 6p 47/83 6p 50/6-4 6p 100/16 6p 100/25 6p 100/25 6p 100/40 6p 100/83 6p 150/16 6p 150/63 6p 220/10 6p 220/15 6p 220/25 6p 220/63 6p 330/10 6p 470/6-3	5p 5p 7p 12p 5p 12p 5p 10p 14p 21p 5p	470/40 500/64 680/6 - 3 680/25 680/40 1000/10 1000/16 1000/25 1000/40 2200/25 2200/40 2800/100 3300/83 3300/100 4700/16	20p 44p 10p 20p 25p 14p 25p 25p 40p 45p 50p 350p 133p 80p	(µF) 0-01 0-015 0-022 0-033 0-047 0-068 0-1 0-15 0-22 0-33 0-47 0-68 1-0 1-2 2-2	3p 3p 3p 3p 3p 4p 5p 7p 1p 14p 24p	(µF/V) 0 · 1/35 0 · 22/35 0 · 22/35 1 · 0/35 1 · 0/35 1 · 0/35 2 · 2/35 4 · 7/35 10/16 10/25 15/6 · 3 22/16 47/16V 100/3	12p 12p 12p 12p 16p 12p 12p 16p 16p 16p 25p 16p
BFY50 BFY52	22p 2N3702 23p 2N3703	12p Z1J	68o P		69p 1A 4		47/25 47/40	8p 470/10 8p 470/25		4700/25 4700/40	75p 93p	PRICES AP		RECT AT TIM	

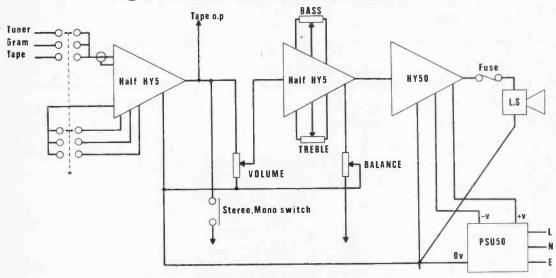
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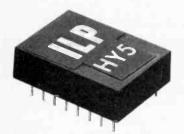
£2 - 40

LLP (Electronics) Ltd

SHEER SIMPLICITY!



MONO ELECTRICAL CIRCUIT DIAGRAM WITH INTERCONNECTIONS FOR STEREO SHOWN



The HY5 is a complete mono hybrid preamplifler, ideally suited for both mono and atereo applications. Internally the device consists of two high quality ampliflers—the first contains frequency equalisation and gain correction, while the second caters for tone control and balance.

TECHNICAL SPECIFICATION
Inputs: Magnetic Pick-up 3mV RiAA: Ceramic Pick-up
3mV. Microphones 10mV. Tuner 100mV. Auxillary 3-100mV:
Input/impedance 47kΩ at 1kHz. Outputs: Tape 100mV.
Main output dot (0.775 RMS). Active Tone Controls:
Treble ± 12db at 10kHz. Bass = 12db at 100Hz. Distortion:
0.5% at 1kHz. Signal/Noise Ratio: 88db. Overload Capabillity: 40db on most sensitive input. Supply Voltage:
±16-25V ± 16-25V

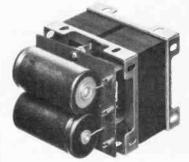


The HY50 is a complete solid state hybrid HI-FI ampifler Incorporating its own high conductivity heatsink hermelically sealed in black epozy resin. Only five connections are provided, input, output, power lines and earth.

TECHNICAL SPECIFICATION Output Power: 25W RMS, into $8k\Omega$. Load Impedance: 4-16k Ω . Input Sensitivity 00b (0.775V RMS). Input Impedance: $47k\Omega$. Distortion: Less than 0.196 at 25W typically 0.05%. Signel/Noise Ratio: Better than 75db. Frequency Response: 104z-56kz± 25V. Size: $105 \times 50 \times 25$ mm.

PRICE £5.98

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The PSU50 can be used for either mono or stereo systems.

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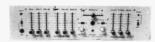
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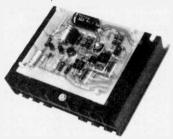
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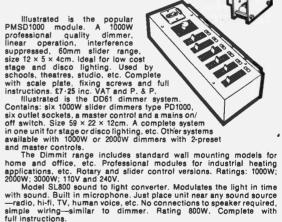


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