

PRACTICAL WIRELESS

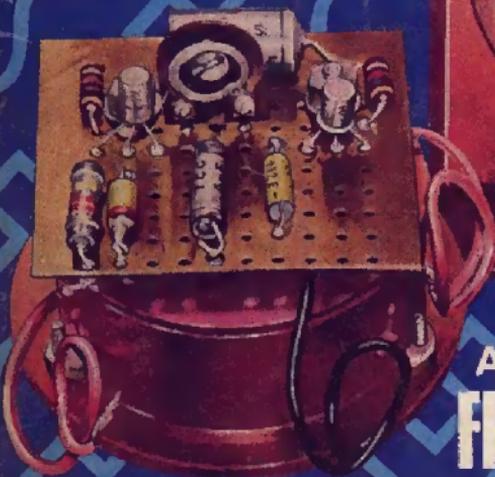
APRIL 1966

2¹

★The PW Reflex-2

A 2-transistor, 4-stage
Pocket Receiver
Simple step-by-step
assembly instructions

★SPECIAL FEATURE ARTICLE



A DIRECT READING
FREQUENCY METER
AND REV. COUNTER

ADCOLA

REGISTERED TRADE MARK
Patent Invention

SOLDERING INSTRUMENTS AND EQUIPMENT



**DESIGNED FOR
THE AMATEUR'S
RADIO STATION**

ILLUSTRATED

List No. 70 $\frac{1}{2}$ " BIT

IN

**PROTECTIVE
SHIELD**

List No. 68

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JACKSON

the big name in PRECISION components

MINIATURE TUNING CAPACITORS TYPE "00"

The Jackson 'O' range contains six basic types of different air-dielectric tuning capacitors with a wide variation of capacities available in each type. In addition, there are optional extras such as concentric-spindle slow-motion drives, built-in trimmers and plastic covers. The maximum capacitance per section ranges from 12, 18 or 24 pF for FM types to 420 pF for AM types.



- ★ Type OO subminiature twin capacitor at 12/6
- ★ Type OFM two-gang for FM at 11/-
- ★ Type OPC for printed circuit mounting at 14/-
- ★ Type OG2 with internal reduction gearing at 16/-

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TRANSISTORISED AUDIO GENERATOR Model 63 £17.1.9

- ★ Laboratory Specification 10 C/s to 100 Kc/s.
- ★ Direct Calibration. ★ Sine and square output



Also available:

- ★ INDUCTANCE BRIDGE 66 ... £18.6.9
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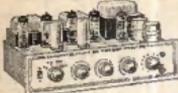
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Estuary House, Camperdown Ter., Exmouth, Devon

R.S.C. STEREO 20 HIGH FIDELITY AMPLIFIER

PROVIDING 10/14 WATT ULTRA LINEAR PUSH-PULL OUTPUT ON EACH CHANNEL



- ★ Four-position tone compensation and Loudness Selector switch.
- ★ Will amplify direct from Tape Deck or Hi-Fi source.
- ★ Stereo/Mono switch.
- ★ Separate Bass "Lift" and "Cut" and Treble "Lift" and "Cut" controls.
- ★ Neon Panel Indicator.
- ★ Handsome Perspex Frontplate.

SUITABLE FOR "MIRK", GRAM, RADIO or TAPE. Employing valves ECC83, ECC85, ECC85, ECC85, ECC85, ECL84, E231. Send S.A.E. for leaflet. FREQUENCY RESPONSE $\pm 3dB$. 30-20,000 c.p.s. HUM LEVEL 46dB down.

SENSITIVITY 23 millivolts, maximum. HARMONIC DISTORTION (each channel) 0.3%. Output transformers are high-quality sectionally wound for specified specification. Output matching for 8 and 16 Ohm speakers. Price £17/8 channel. Complete set of parts with point-to-point wiring diagrams and instructions, or factory assembled, tested and supplied with our usual 12 months' guarantee for

13 Gns.

18 Carr. 12/6 or DEPOSIT 6/7 and 9 monthly GNS. payments 3/10 (Total £20.15.6).

R.S.C. A15 20 WATT R.M.S.

HIGH FIDELITY TRANSISTOR AMPLIFIER

With integral pre-amp tone control stages



Output for 8, 16 and 15 ohm speakers. Kit of parts consisting of Printed Circuit and all components for same including 9 Mullard or Newmarket latest type semi-conductors. Includes neat full wiring instructions or with printed circuit fully wired and tested £2 extra. Post 5/-

£6.19.9

Frequency Response $\pm 1dB$ 20-20,000 c.p.s. Harmonic Distortion 0.1% measured at 1000 c.p.s. Hum and Noise $\pm 80dB$. Sensitivity 20V. Bass Control $\pm 6dB$ to $-12dB$ at 40 cps. Treble Control $\pm 4dB$ to $-12dB$ at 10 Kcs. Suitable Power Pack Kit 3/9/6, Or ready built 5/9/6

AUDIOTRINE HI-FI TAPE RECORDER KIT

REALISED AT INCREDIBLY LOW COST, CAN BE ASSEMBLED IN AN HOUR. ONLY 4 PAIRS OF SOLDERED JOINTS PLUS MAINS. Incorporating the latest Magnavox Tape Deck. High Quality Tape Amplifier with equalisation for each of 3 speeds. 3 1/2" Flux P.M. Spinning empty Tape Spool, a Reel of Best Quality Tape and a Handsome Portable Carrying Cabinet of latest styling and finished dark grey leathercloth with silver handles. A specially designed 14 x 17 x 9 1/2 in. high and circuit. Total cost if purchased individually approx. £35. Performance equal to units in the £50-£80 class. S.A.E. for leaflets. TERMS: Deposit 4/8, and 12 monthly payments of 4/2/- (Total 28 Gns.). 4 Track Model 3 Gns. extra.



25 1/2 Gns.

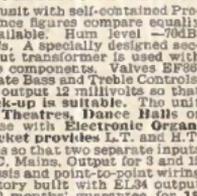
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ON H.P. ACCOUNTS SETTLED IN 6 MONTHS

LINEAR TAPE PRE-AMPLIFIER. Type LP.1. Switched Equalization. Positions for Recording at 14 in., 10 in., 7 1/2 in. per sec. and Playback E.M.S. Recording Level Indicator. Designed primarily as the link between a Magnavox Tape Deck and Hi-Fi amplifier suitable almost any Hi-Fi Tape Deck. S.A.E. for leaflet. **9 1/2 Gns.**

R.S.C. A10 30 WATT ULTRA LINEAR HIGH FIDELITY AMPLIFIER

A highly sensitive Push-Pull high output unit with self-contained Pre-amp./Tone Control Stages. Performance figures compare equally with most expensive amplifiers available. Hum level $\pm 70dB$. Frequency response $\pm 3dB$ 30-20,000 c.p.s. A specially designed sectionally wound ultra linear output transformer is used with 80V output valves. All first grade components. Valves EF86, EF87, ECC83, 6BT, 6Z3. Separate Bass and Treble Controls. Minimum input required for full output 12 millivolts so that any kind of Microphone or Pick-up is suitable. The unit is designed for Clubs, Schools, Theatres, Dance Halls or Outdoor Functions. Complete kit of parts with point-to-point wiring diagrams and instructions. **12 Gns.** Carr. Supplied factory built with EL84 output valves. 10/-



G.S. If required perforated cover with carrying handles can be supplied for £1. TERMS: Deposit 4/8 and 9 monthly payments of 3/3 (Total £17.3.6).

HIGH FIDELITY 12-14 WATT AMPLIFIER TYPE A11

PUSH-PULL ULTRA LINEAR OUTPUT "BUILT-IN" TONE CONTROL PRE-AMP

Two input sockets with associated controls allow mixing of "mike" and gram. etc. High sensitivity. Valves ECC83, ECC83, EL84, EL84, E231. High quality sectionally wound output transformer specially designed for Ultra Linear operation and reliable small condensers, etc. INDIVIDUAL CONTROLS FOR BASS AND TREBLE "Lift" and "Cut". Frequency response $\pm 3dB$ 30-20,000 c.p.s. Six negative feedback loops. Hum level $\pm 60dB$. SENSITIVITY 23 millivolts. Suitable for Crystal or Ceramic P.U.'s all types "mikes". Comparable with the very best designs. For Musical Instruments such as String Bass, Electronic Guitars, etc. Output Socket provides 30V, 50V, and 80V. L.S. for supply of a Radio Tuner. Size approx. 12 x 10 x 7 1/2 in. For A.C. mains 200-250V, 50 c.p.s. Output for 3 and 15 ohm speakers. Kit complete to last unit. Chassis fully punched. Full instructions and point-to-point wiring diagrams supplied. Only **8 Gns.** (Or factory built 5/1/8 extra).



Metal covers with 2 carrying handles can be supplied for £1. TERMS ON ASSEMBLED UNITS: Deposit 3/3/8 and 9 monthly payments of 3/4 (Total £12.16.0). Send S.A.E. for illustrated leaflet detailing Cabinets, Speakers, etc.

R.S.C. STEREO TEN HIGH QUALITY AMPLIFIER

A complete set of parts for the construction of a stereo amplifier giving 5 watts high quality output on each channel (total 10 watts). Sensitivity is 50 millivolts. Suitable all crystal or ceramic stereo heads. Ganged Bass and Treble Control give equal variation of "lift" and "cut". Provision is made for use as sprinter (mono) and 10 watt amplifier. Valve line-up ECC83, ECC83, EL84, EL84, E231. Outputs for 3-ohm speakers. Point-to-point wiring diagrams and instructions supplied. Send S.A.E. for leaflet. **8 Gns.**



Or supplied factory assembled with 12 months' guarantee for £11.7.6. Carr. 10/-

TERMS: Deposit 2 Gns. and 9 monthly payments of 2/4 (Total 21.15.6).

R.S.C. 4/5 WATT AS HIGH GAIN AMPLIFIER

A highly-sensitive 4-valve quality amplifier for the home small club, etc. Suitable for all crystal or ceramic P.U. head units. Practically all "mikes". Separate Bass and Treble controls giving "lift" and "cut". Hum level 71dB down. Negative Feedback 15dB. H.T. of 300V. 55mA and wiring diagrams supplied. L.S. for supply of a Radio Tuner or Tape Deck Pre-amp. For A.C. mains 200-250V. Speaker output 5 ohms. Kit is complete in every detail with fully illustrated Handsome finished chassis, point-to-point wiring diagrams and instructions. Exceptional value **£4.15.0** or assembled ready for use 2/5/- extra, plus 3/6 carr. deposit 2/2/6 and 5 monthly payments of 2/8 (Total £2.15.6), for assembled unit.

CORNER CONSOLE CABINETS

Strongly made. Beautiful polished walnut veneered finish. Pleasing design. JUNIOR MODEL. For up to 2 in. speaker. Approx. 20 x 11 4/9 x 3 in. Carr. 5/6. STANDARD MODEL. To take up to 10 in. speaker. Size 37 x 18 x 10 in. **5 Gns.** Carr. 7/6. SENIOR MODEL. To take up to 12 in. speaker and with Tweeter cabinet. Size approx. 30 x 30 x 15 in. (Recommended for use with Audiotrine speaker system.) Carr. 8/6. Terms available.



12 in. HIGH QUALITY LOUSPEAKERS

In walnut veneered cabinet. 10 WATT MODEL. Class A, 12,000 lines, speech cut 3 or 15 ohms. **£4.19.6** Carr. 5/6. Terms: Deposit 1/5/- and 9 monthly payments of 10/10 (Total £5.12.6).



30 WATT MODEL. 15 ohms. Size 15 x 15 x 10 in. Terms: Deposit 2/6/- and 9 monthly payments of 17/- (Total £8.17.6). Carr. 8/6. 80 WATT MODEL. 15 ohms. Carr. 10/- or Deposit 3/5/- and 9 monthly payments of 22/- (Total £11.15.0). Any of above in extra heavy rexine covered Cabinets, 21 extra.

R.S.C. JUNIOR BASS REFLEX CABINET. Designed to take 15 ohm high quality speaker. Acoustically lined and ported. Polished Walnut veneer finish. Size 18 x 12 x 10 in. Strongly made. Handsome appearance. Superb reproduction. Carriage 5/-. **£4.7.6**

AUDIOTRINE HI-FI SPEAKER SYSTEMS

Consisting of matched 12 in. 12,000 lines 15 ohm high quality speaker; cross-over unit (consisting of choke, condenser, etc.) and Tweeter. Smooth response and extended frequency range ensure surprisingly realistic reproduction. Standard 10 x 7 1/2 in. size. Carr. **£4.19.9** Or Senior 20 watt, £6.19.6 Carr. 7/6. Or Deposit 2/1/- and 8 monthly payments of 1/5/- (Total £7.16.0).



TWEETERS

Fane 3 ohm or 15 ohm 25/9.

R.S.C. BABY ALARM or INTERCOM KIT. All parts with diagrams etc. In two polished walnut finished cabinets of pleasing design. High sensitivity. For 200-250V, 50 c.p.s. Fully isolated. Controllable at both units. A unit of this class would normally cost £20-£30. Only 5/9/6, carr. 5/-. Ready for use. 6 Gns.

R.S.C. BASS-REGENT 50 WATT AMPLIFIER

AN EXCEPTIONALLY POWERFUL HIGH QUALITY ALL-PURPOSE UNIT

For lead, rhythm, bass guitar and all other musical instruments
For vocalists, gram, radio, tape and general public address



* UNUSUALLY POWERFUL LOUDSPEAKER COMBINATION consisting of a FANE HIGH FLUX 15in. 30 watt unit PLUS a FANE 12in. 30 watt unit with extended frequency response. 4 Jack Inputs and two Volume Controls for simultaneous use of up to 4 pick-ups or "mikes".

* Cabinets covered in two-tone Rexine/Vynair with gold trimmings.

* Separate Bass and Treble Controls giving 'lift' and 'cut'. Send S.A.E. for leaflet. Or call at one of our branches and compare with units at three times the cost.

49 GNS. Or deposit of £7.18.0 and 12 monthly payments of Carr. 25/- 81/6 (Total 54 5/6s.).

R.S.C. B20 MULTI-PURPOSE AMP. especially suitable for Bass Guitar

With Massive 15in. high flux loudspeaker. Rating 25 watts. Individual bass and treble controls. Two jack inputs separately controlled. Heavy cabinet attractively finished in Rexine/Vynair. Size approx. 8 1/2 x 11 x 4 1/2 in. Send S.A.E. for leaflet. Carr. 17/6 Or deposit 24.14.6 and 12 monthly payments of 49/- (Total 59.14.6).



R.S.C. G15 IS WATT AMPLIFIER for Lead or Rhythm Guitar, Mike, Gram or Radio

High-fidelity output. Separate bass and treble controls. Twin separately controlled inputs. Heavy Duty 12in. 50 watt speaker. Cabinet covered in attractive Rexine/Vynair. Size approx. 18in. x 10in. x 4 1/2 in. Carr. 12/6. Or Deposit 8 gns. and 12 monthly payments of 32/6 (Total 22.15.6).

JASON FM/L VHF/FM Radio Tuner design. All parts inc. valves. £6.19.6

TANNOY RE-ENTRANT LOUD-SPEAKERS. For outdoor or 27/6 (Carr. 5 gns. 3 watts) 4/6

R.S.C. MAINS TRANSFORMERS (FULLY GUARANTEED)

Interleaved and Impregnated. Primaries 250-250-250 v. 50 c/s. Screened. MIDGET Clamped Type 24 x 24 x 2 1/2 in. 60ma. 6.3v. 2a. 14/9

250-250V. 60ma. 6.3v. 2a. 14/9

250-250V. 60ma. 6.3v. 2a. 14/9

250-250V. 100ma. 6.3v. 2a. 17/9

TRANSISTORISED SOUND MIXER

Enables mixing of up to 4 standard jack inputs, i.e. mic, tape, gram, tuner, etc., into single output. Compact and completely self-contained, uses standard 9v. battery. **49/9**

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ACCOUNTS SETTLED IN

6 MONTHS



(MANCHESTER) LTD.

Addresses Page 1010

R.S.C. COLUMN SPEAKERS

Covered in two-tone Rexine/Vynair. Ideal for vocalists and Public Address. Normally supplied for 15 ohm matching but can be supplied for 100v. line for 30/- extra.

Type C38, 15-30 watts. Fitted

5in. high flux speakers. Overall

size 5 1/2 x 4 1/2 x 4 1/2 in.

12 1/2 Gns. Carr. Or deposit

3 monthly payments of 27/9

(Total 51.4.9).

Type C42, 40 watts. Fitted

40 12in. 12,000 line 10 watt

speakers. Overall size 5 1/2 x 4 1/2

5in. approx. 19 Gns.

Carr. 15/-

Or deposit 3 gns. and 3 monthly

payments of 43/6 (Total 51.6s.).

30 WATT HI-FI AMPLIFIER

For ELECTRONIC ORGAN, GUITAR, VOCAL or INSTRUMENTAL GROUPS



A Four Input two volume control hi-fi unit with separate Bass and Treble 'cut' and 'boost' controls. Attractive 30 watt valve. Strong Rexine covered cabinet. Attractive Black/Gold fascia plates. For 200-

250V. A.C. mains. Output for 3 or 15 ohm

speakers. Send S.A.E. for leaflet.

Carr. 12/6 Or Deposit 3 gns. and 3 monthly

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17 Gns. £19.13.0).

18in. 60 WATT EXTRA HEAVY DUTY LOUDSPEAKERS

Famous Carr. 15/-

each. Normal price over 17 Gns.

22S. Very limited number

to clear fully guaranteed. Terms available.

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0A2	8/0	3BX6	4/8	7B2	9/8	3Z5Z	7/8	AC/FPN 4/9	8AF4Z	7/9	EL84	4/6	L2319	6/8	ROL340A	U901	11/-	GD14	10/-	
0A2	8/0	3BX6	4/8	7B2	9/8	3Z5Z	7/8	AC/FPN 4/9	8AF4Z	7/9	EL84	4/6	L2319	6/8	ROL340A	U901	11/-	GD14	10/-	
0C24GT	4/8	3C0	2/5	7C3	6/8	28D1	8/9	(b) 1/78	EB4	1/-	EL83	7/6	LK329	6/8	8K34	65/-	U839	9/-	GT103 7/-	
1A3	2/6	8C8	3/-	7C7	5/-	39C1	6/8	AC/FPN	EB91	2/8	EL91	2/8	MHL4D12B	6/8	S135	22/6	U839	9/-	GT104 10/-	
1A4	1/6	3C	2/-	7D3	1/5	39C15	10/-	(b) 1/77	EB93	3/6	EL92	6/8	MHL4D12B	6/8	S135	22/6	U839	9/-	GT104 10/-	
1A5	5/-	6C9	10/10	7D6	14/6	39C17	11/8	AC/8G 28/6	EB93	3/6	EL90	8/7-	M84D	6/8	SP13C	18/6	U91	15/-	GT113 6/8	
1A7GT	7/6	8C10	8/-	7F7	7/-	39C18	9/3	AC/8G1V	ED04	6/8	EL90	15/4	M84D	12/-	SP41	2/-	U4929	8/6	GT114 6/8	
1C1	4/-	8C13	5/8	7F7	5/9	39F3	7/3	AC/8G1V	EL95	2/6	EL95	22/6	P12214	6/8	SP42	18/6	VMF43	12/0	GT117 7/-	
1C2	6/6	8C16	18/-	7F7	15/6	39F11	8/2	AC/8G1V	EL95	2/6	EL95	22/6	P12214	6/8	SP42	18/6	VMF43	12/0	GT117 7/-	
1C3	6/6	8C17	9/6	7X	5/-	39F11A	11/8	AC/TP 18/-	EB91	5/8	EM85	12/-	N8	10/8	SV25	27/2	YP2	8/6	GT174 8/8	
1C5	5/-	8C18	8/6	7Y4	5/-	39L1	8/6	AC/VP1 18/-	EBF90	5/9	EM7	12/8	N78	22/6	TA1	9/-	VP28	9/6	GT183 8/6	
1C6	10/6	8C19	24/-	8D2	2/4	39L15	10/6	AC/VP2 12/6	EB95	7/6	EM8	6/8	N108	20/2	TBD2	12/6	VM4	14/8	GT184 8/6	
1D6	9/6	8D3	9/6	9D2	3/-	39L17	11/6	ATP4	EMF99	9/9	EMM1	7/-	N39	25/2	TD4	7/6	VP4	14/6	GT186 10/-	
1ED1	4/6	8D6	3/-	9D7	7/6	39P13	7/8	AZ1	EM84	4/6	EN3	10/6	PC8	9/6	TR4	10/-	VP4	14/6	GT186 10/-	
1ED2	9/6	8D7	9/6	9D7	7/6	39P15	8/6	AZ1	EM84	4/6	EN3	10/6	PC8	9/6	TR4	10/-	VP4	14/6	GT186 10/-	
1G6	6/6	8F1	9/6	10K2	12/-	39P16	9/6	B38	4/9	EC22	9/8	EN1	5/6	PC8	9/6	TR4	10/-	VP4	14/6	GT186 10/-
1HE2GT	7/6	8F6	3/6	10D1	7/-	39P17	9/6	8B49	10/8	EC70	4/9	EN1	5/6	PC8	9/6	TR4	10/-	VP4	14/6	GT186 10/-
1H4	2/8	8F8	9/6	10D2	11/8	39P18	10/8	8T18	5/8	ECF90	7/8	ET51	6/8	PC8	9/6	TR4	10/-	VP4	14/6	GT186 10/-
1L4A	17/6	8F7G	8/-	10F1	10/-	39P14A	11/8	8L43	10/8	EM1	4/6	ET4	7/2	PC97	6/8	TP25	5/-	VR103	5/6	GT184 6/8
1L4B	16/10	8F8	5/-	10F3	9/6	39P15	9/6	C	10/8	EC92	6/8	EV48	9/3	PC84	8/6	TP2690	7/6	VR190	4/8	GT184 6/8
1LD6	4/6	8F12	3/6	10P18	9/9	39P18	12/6	CK006	6/6	ECU31	7/8	EV54	9/3	PC86	9/6	TY98	11/8	VR184	7/8	GT184 6/8
1LNGT	8/6	8F13	9/6	10P19	9/9	39S48	8/9	CV1	10/8	EC45	9/6	EV54	9/3	PC86	9/6	TY98	11/8	VR184	7/8	GT184 6/8
1N1	6/6	8F14	25/-	10D1D	9/6	39S49	8/9	CV8	11/8	EC53	9/1	EV57	8/7	PC88	11/6	UB43	7/6	VU11	6/6	GT184 6/8
1P1	6/6	8F15	6/6	10P13	12/-	39S47	4/6	CV8	2/6	EC54	21/7	EV57	8/7	PC88	11/6	UB43	7/6	VU11	6/6	GT184 6/8
1P10	4/6	8F16	6/6	10P14	11/6	39Z42	10/2	CV21	10/6	EC55	4/8	EV51	8/7	PC88	11/6	UB43	7/6	VU11	6/6	GT184 6/8
1P11	4/6	8F17	12/6	11D5	2/6	39Z43	4/6	CV21	10/6	EC55	4/8	EV51	8/7	PC88	11/6	UB43	7/6	VU11	6/6	GT184 6/8
1R4	4/6	8F18	18/6	11D6	17/6	39Z44	4/6	CV28	18/6	EC58	3/1	EV3	8/9	PC88	11/6	UB43	7/6	VU11	6/6	GT184 6/8
1R4	4/6	8F18	18/6	11D6	17/6	39Z44	4/6	CV28	18/6	EC58	3/1	EV3	8/9	PC88	11/6	UB43	7/6	VU11	6/6	GT184 6/8
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1R4	4/6	8F18	18/6	11D6	17/6	39Z44	4/6	CV28	18/6	EC5										



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3" PORTABLE 'SCOPE. OS-2 With improved performance and modern styling. "Y" bandwidth 2 c/s-3 mc/s. Automatic Lock-in sync. Mumetal C.R.T. shield. P.C. Board. Size 5" x 7" x 12" deep. Weight 9½ lb.
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5in. OSCILLOSCOPE Model 10-12U. Laboratory quality at utility oscilloscope price. Wide band amplifiers essential for T.V. servicing. F.M. alignment, etc. T/B covers 10 c/s-500 kc/s in 5 ranges.
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REGULATED POWER SUPPLY Model IP-20U. Transistorised, 0.5-50 v. D.C. Up to 1.5 amps. Compact: 9½ x 6½ x 11 in.
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4½" VALVE VOLTMETER, Model V-7A. The world's best selling VTYM. Measures up to 1,500 volts (d.c. and r.m.s.) and 4,000 pk. to pk. Res. 0.1 ± 1,000 Mc/s. Centre zero dB scale, d.c. input resistance 11MΩ. 4½in. meter. Complete with test prods, leads and standardising battery.
£19.18.6 Assembled **£13.18.6** Kit



V-7A

DE-LUXE 6in. VALVE VOLTMETER, Model IM-13U. Similar spec. to model V-7A but with improved accuracy. Larger meter. Unique gimbal mount.
£26.18.0 Assembled **£18.18.0** Kit



IG-82U

SINE WAVE GENERATOR, Model IG-82U. Freq. range 20 c/s-1 Mc/s in 5 bands. Simultaneous sine and square wave outputs.
£36.10.0 Assembled **£24.10.0** Kit

HARMONIC DISTORTION METER, Model IM-12U. Freq. cov.: 20-20,000 c/s.
£34.0.0 Assembled **£24.15.0** Kit

TV ALIGNMENT GENERATOR, Model HFV-1. Covers 3.6 to 220 Mc/s fundamentals.
£47.10.0 Assembled **£37.18.0** Kit

RF SIGNAL GENERATOR, Model RF-1U. Up to 100 Mc/s fundamental, 200 Mc/s harmonics. Up to 100 mV output on all bands.
£19.18.0 Assembled **£13.8.0** Kit

MULTIMETER, Model MM-1U. Ranges: 0-1.5 v. to 1,500 v. a.c. and d.c.; 150µA to 15A d.c.; 0.2Ω to 20MΩ. 4½in. 50µA meter.
£18.11.6 Assembled **£12.18.0** Kit

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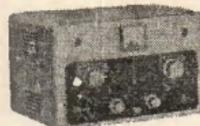
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Concurrent with the Audio Fair.

The Grand Hotel, Southampton Row, London.
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AMATEUR EQUIPMENT

AMATEUR TRANSMITTER, Model DX-100U. Covers all amateur bands 160-10M. 150 w. d.c. input, self contained with power supply. Modulator, VFO
Assembled **£104.15.0** **£79.10.0** Kit



DX-40U

AMATEUR BANDS RECEIVER, RA-1 Outstanding performance at low cost. Covers the amateur bands from 160-10 metres. Half lattice filter. Signal strength meter.
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AMATEUR TRANSMITTER Model DX-40U. Covers 80-10 m. Power inputs 75 w. C.W., 60 w. peak C.E. phone. Output 40 w. to aerial. Prov. for VFO.
£33.19.0 Kit
Assembled **£45.8.0**



Other kits in the amateur range include: SSB Adaptor SB-10U £39.5.0. Variable freq. Oscillator, VF-1U £10.17.6. Balun Coil Units B-1U £4.15.6. Grid-Dip Meter GD-1U £10.19.6. Q Multiplier QPM-1 £8.10.0. Reflected Power Meter HM-1U £8.5.0. Receiver, RG, £39.16.0.

SPEAKER SYSTEMS

BERKELEY Slim-line LOUDSPEAKER SYSTEM

★ Now from Heathkit, a loudspeaker kit with a fully finished walnut cabinet. ★ A 2 speaker system. ★ Modern, slim-line styling. ★ 30 c/s to 17 kc/s. ★ Only 7½" thick. ★ Use it horizontally or vertically. ★ Stand it on the floor or in your bookshelf.
£23.0.0 Assembled **£18.10.0** Kit



Berkeley

COTSWOLD STANDARD MODEL

Acoustically designed enclosure "in the white" 26 x 23 x 15½in., 12in. bass speaker, elliptical middle speaker, 2in. pressure unit. Covers 20-20,000 c/s. Complete kit with all controls.
£25.12.0 Kit



MFS SYSTEM (illustrated)

A minimum floor space model for the smaller room. 26in. high x 16½in. x 14in. deep. Similar performance to standard model.
£25.12.0 Kit

SSU-1 SYSTEM

A practical solution to the problem of a moderately priced speaker suitable for Stereo Mono amplifiers where the equipment has to be compact. Two speakers, balance control, ducted port reflex cabinet.
Incl. P.T. **£11.7.6** Kit
(less legs)



Set of 4 legs (7" or 15") 14.6 extra.

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MODELS FOR HOME, TEST AND WORKSHOP.
BRITISH HEATHKIT MODELS USE BRITISH COMPONENTS



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DE-LUXE ALL TRANSISTOR STEREO AMPLIFIER AA-22U.

A quality British transistor amplifier with high power (20 + 20W) at a reasonable cost, capable of delivering full power at all frequencies in the audio band. Handsome fully finished walnut veneered cabinet available. Professional slim-line styling. Send for full specification. Kit (less cabinet) £39.10.0 Cabinet £25.0. Accessories for panel mounting £1.18.0.



6W DE-LUXE STEREO AMPLIFIER. Model S-33H. An inexpensive stereo amplifier with high sensitivity. Suitable for use with Decca Derram cartridge.

£21.7.6 Assembled **£15.17.6** Kit

TAPE RECORD/REPLAY AMPLIFIER KITS. Will operate with most tape decks. Send for details.

TA-1M (Mono), 19.18.0 Kit. TA-1S (Stereo), £25.10.0 Kit.

STEREO CONTROL UNIT. Model USC-1. Ideal for use with MA-12 amplifiers. Kit £19.10.0. Assembled £26.10.0.



S-33

6W STEREO AMPLIFIER

Model S-33. 3 w/ch. Inputs for radio, tape and gram. Stereo/Mono range controls. Sensitivity 200mV. £18.18.0 Assembled **£13.7.6** Kit

TRANSISTOR MIXER. Model TM-1. A must for the Tape enthusiast. 4 channels, Battery driven.

£16.17.6 Assembled **£11.16.6** Kit

5W HI-FI MONO AMPLIFIER. Model MA-5. A low priced amplifier based on the S-33. Printed circuit construction makes it easy to build.

£15.10.0 Assembled **£10.19.6** Kit

HI-FI MONO POWER AMPLIFIER. Model MA-12. Ideal for use with Models USC-1 and UMC-1. 0.1 THD at 10W. Wide frequency range.

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18W STEREO AMPLIFIER.

Model S-99. Ganged controls.

Stereo/Mono gram, radio and tape

record inputs. PiB selection.

£38.9.6 Assembled **£28.9.6** Kit



S-99

A range of R/players, T/Decks available.
Send for full spec. of any model

Prices quoted are Mail Order and include free delivery in UK.

TRANSISTOR RECEIVERS

"OXFORD" LUXURY TRANSISTOR DUAL WAVEBAND RECEIVER.

The ideal domestic or personal portable receiver. 10 Semi-conductors. Soft leather case. Send for full details.

Incl. P.T. **£14.18.0** Kit



6 TRANSISTOR PORTABLE. Model LXR-1. Presigned I.F. transformers. Printed circuit, 7in. x 4in. high flux speaker. Real hide case. Very easy to build.

Incl. P.T. **£12.11.0** Kit



GC-1U

"MOHICAN" GENERAL COVERAGE RECEIVER. Model GC-1U. Excellent portable or general purpose receiver for "amateur" or short wave listening. See full spec. leaflets.

Assembled £45.17.6 **£37.17.6** Kit

PA/GUITAR AMPLIFIERS

Model PA-1 (similar to PA-2 illustration) 50 watt output. Two 12" L.S. Variable tremolo, 4 inputs.

Assembled £74 **£54.15.0** Kit

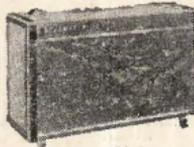
Legs or castors available (extras). Model PA-2 (illustrated).

Starmaker 33. All transistor 20 watt output, two 12" L.S. Variable tremolo, 4 inputs.

Compact size. Assembled £59.10.0 **£44.19.0** Kit

Legs or castors available (extras). 50 watt Power Amp. MA-50

Assembled £27.18.0 **£19.18.0** Kit



PA-2

HI-FI TUNERS

Model FM-4U. Tuning range 88-108 Mc/s. Tuning unit (FM-T-4U) with 10.7 Mc/s. I.F. (£2.15.0 incl. P.T.). I.F. Amp. (FMA-4U) complete with cabinet and valves (£13.13.0). Total **£16.8.0** Kit

Assembly can be arranged.

AM-FM TUNER. Covers FM 88-108 Mc/s. A.M. 16-50, 200-550, 900-2,000 m. Tuning heart (£4.13.6 incl. P.T.), and I.F. Amp. (£22.11.6). Total **£27.5.0** Kit

Send for leaflets. Assembly can be arranged.



FM-4U



EQUIPMENT CABINETS

A large range, in kit form or assembled and finished, available to meet most needs. Illustrated details on request.

Prices from **£8.8.0** to **£46.4.0**

SPEAKERS FOR YOUR OWN ENCLOSURE

12" Heavy-duty Bass (Fane 12/12) £7.7.0.

2" Tweeter (Fane 301) £3.1.6.

(both as used in the Cotswold systems).

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Two Speakers + Cross-over, System SCM-1.

(As used in model SSU-1) with details for enclosure £5.5.0.

ELECTRONIC WORKSHOP KIT. Model EW-1. An outstanding experimental kit for children. £7.13.6 incl. P.T.

A WIDE RANGE OF BOOKS ON ELECTRONICS AND RADIO. PLEASE SEND FOR LISTS AND PRICES.

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We have opened a showroom and retail stores at 233 Tottenham Court Road.

WHEN YOU ARE IN TOWN, WE HOPE
THAT YOU WILL VISIT US THERE

To DAYSTROM LTD. Dept. P.W.-4, GLOUCESTER, ENG.

Please send me FREE BRITISH CATALOGUE (Yes/No) Full details of model(s). Amateur Brochure (Yes/No), American Catalogue 1/- (Yes/No), Instrument Brochure (Yes/No)

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ADDRESS _____

LASKY'S RADIO

Offer the Finest Value and HOME CONSTRUCTORS

We consider our construction parcels to be the finest value on the home constructor market. If on receipt you feel not competent to build the set, you may return it as received within 7 days when the sum paid will be refunded less postage.

THE "SKYROVER" RANGE

GENERAL SPECIFICATION

7 transistor plus 2 diode superhet, 6 waveband portable receiver. Operating from four 1.5 v. torch batteries. The SKYROVER and SKYROVER DE LUXE cover the full Medium Waveband and short Waveband 31-64 M. and also 4 separate switched band-spread ranges, 15M., 16M., 18M. and 23M., with Band Spread Tuning for accurate Station Selection. The coil pack and tuning heart is completely factory assembled, wired and tested. The remaining assembly can be completed in under three hours from our easy to follow stage by stage instructions.

A simple additional circuit provides coverage of the 1100/1550M. band (including 1500 M. Light programmes). All components and detailed construction data.

Only 10/- extra Post Free
This conversion is suitable for both models that have already been constructed.

Data for Receiver 276 extra. Refunded if you purchase the parcel. Four U2 batteries, 3/4 extra. All components available separately.

NEW—The SKYROVER Mk 111

New supplied with redesigned cabinet, edgewise controls, black and chrome plastic cabinet. Size 10 x 6 1/2 x 3 1/2 in. with carrying handle. Can now be built for H.P. Terms: 27/- dep. and 11 months at 15/6. Total H.P. Price £10-0-5.

NEW—The SKYROVER DE LUXE

Tune Control Circuit is incorporated with separate Control. In a wood cabinet, size 11 1/2 x 6 1/2 x 3 1/2 in., covered with a washable material with plastic trim and carrying handle. Car aerial socket fitted. Can now be built for H.P. Terms: 33/- dep. and 11 mths. at 19/6. Total H.P.P. £12-3-10



The Very Latest MAGNAVOX-COLLARO 363 TAPE DECKS

3 speed model—i.e. 31, 7 1/2 i.p.s., available with either 1 track or 2 track heads. Features include: pause control; digital counter; fast forward and rewind; Bow 4 pole fully screened induction motor; interlocking keys. Size of top plate 13 1/2 x 11 x 5 1/2 in. deep below unit plate. For 200/250V. A.C. mains, 50 c.p.s. operation. New, unused and fully guaranteed.

LASKY'S PRICE with 1 track heads **£10.10.0** Carriage and packing 7/6 extra
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SPECIAL for OVERSEAS CUSTOMERS—the new Magnavox-Collaro 363 Deck for 110/125V. 50 or 60 c.p.s. mains now available, prices as above. Post to any part of the World, 35/-

SPECIAL PURCHASE! 7 TRANSISTOR RADIO CHASSIS

Superhet chassis, fully built—by famous British manufacturer—on printed circuit. Uses 7 Newmarket transistors giving 500 mV push-pull output—5 L.P. stages; ferrite rod aerial. Covers full, long and medium wavebands—300-3000M and 1200-3000M and switched band spread on 2043 (Luxembourg), 41m, speaker. Steel; volume control on/off switch, tuning dial and position for car aerial socket. Uses any 9V battery. Overall dimensions—7in. x 4 1/2 in. x 2 1/2 in. Absolutely complete except for cabinets.

LASKY'S PRICE 69/6 Post

Recommended battery PP6, 2/6 or PP7, 3/3 extra.

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GET S1, GET S5, GET S8, 3/6; 371A, 374F, 3/6; OC46, OC71, OC81D, 4/6; OC44, OC70, OC75, OC81 (match pair 10/6), 5/6; AF17, OC70, OC200, 8/6; OC28, OC43, OC72, 7/6; OC201, OC204, 15/-; OC200, OC206, 15/6; OC28 24/6.

TRANSISTERS by BRUSH CRYSTAL 60

TO-018 465 kva., 2 kva., 7/6 EACH
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TO-02B 465 kva., 1 kva., 7/6 EACH
TO-02D 470 kva., 1 kva., 7/6 EACH
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B.S.R. AUTOCHANGERS—BRAND NEW AT LOWEST EVER PRICES

Brand new and fully guaranteed—complete with cartridge and stylus.

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Add 5/- carriage and packing on each.

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SPECIAL OFFER: GARRARD A75 AUTOCHANGERS

4 speed mains model—brand new and fully guaranteed
LASKY'S PRICE With mono crystal cartridge **£5.19.6** With stereo cartridge **£6.9.6** Post 6/-

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The "REALISTIC" Seven

★ 7-Transistor Superhet. ★ 350 milliwatt output into 4in. high imp. speaker.

★ All components mounted on a single printed circuit board. ★ Full modulation and long wave cover. ★ Plastic cabinet with carrying handle, size 7 x 10 x 3 1/2 in. Blue Grey or all Grey.

★ Socket for car aerial. ★ Frequency 470 Kcs. ★ Ferrite rod internal aerial. Operates from PP6 or similar battery. ★ Full data supplied.

All coils and I.F.'s etc. fully wound ready for immediate assembly. An outstanding receiver for **£5.19.6**

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3in. Triple play, 800ft., Mylar base	15 0
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8in. Double play, 1,200ft., Mylar base	16 0
8in. Long play, 900ft., Acetate base	10 0
8in. Standard play, 900ft., PVC base	8 6
8in. Triple play, 1,200ft., Mylar base	25 0
5 1/2in. Double play, 1,800ft., Mylar base	22 6
5 1/2in. Long play, 1,200ft., Acetate base	18 6
5 1/2in. Standard play, 900ft., PVC base	11 6
5 1/2in. Standard play, 1,200ft., Mylar base	15 0
7in. Triple play, 2,400ft., Mylar base	45 0
7in. Standard play, 1,800ft., Acetate base	10 0
7in. Standard play, 1,200ft., Mylar base	15 0
7in. Long play, 1,800ft., Mylar base	18 6
7in. Double play, 2,400ft., Mylar base	26 0
7in. Long play, 1,800ft., Acetate base	15 0
7in. Triple play, 2,400ft., Mylar base	68 0
P. & P. 1 1/2-extra per reel, 4 reels and over Post Free.	

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We stock the complete range of Sinclair Super-miniature Kits.
THE MICRA-20 pocket radio 59/6
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THE X-50 Amplifier £7.19.6
Ready built and tested 25.19.6
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See our Practical Wireless November '66 advertisement for details.

U.H.F. TUNERS

Only a few available. Complete with PCB and PCB5 valves. We regret no circuit or data is available. Knobs included.

LASKY'S PRICE 59/6

Post 2/6.

GORLER UT 340 FM/VHF TUNING HEART

Permeability tunes—covering 87 to 108 Mc/s. For use with one ECC85 valve. In metal case, size 3 1/2 x 4 1/2 in. Circuit supplied.

LASKY'S PRICE 15/11

ECC85 valve 9/- extra. Post 2/-

Service in Great Britain to both & HI-FI ENTHUSIASTS

TAPE RECORDERS · RECORD PLAYERS · AMPLIFIERS, ETC. COMPLETE MONO/STEREO SYSTEMS TO YOUR SPEC.



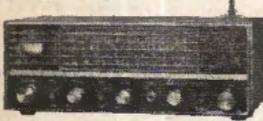
NEW COMMUNICATION RECEIVERS MODEL KT 320 KIT



Designed in sub-assemblies for easy building. Covers range from 540 Kc/s to 30 Mc/s. Ham Band is provided with a scale for direct reading and can also be band spread. 8 valves. Facilitator: A.N.L., A.V.C. and M.V.C. Q-Multiplier also serves as B.F.O. H.F. stage and two I.F. stages ensure high sensitivity and selectivity (all coils and L.P.s are supplied pre-aligned). 2 Aerial Sockets. Stand-by position for use with a transmitter. 8 meter fitted. 200-250V. A.C. mains. Steel cabinet, grey crackle finish. Size 15 x 8 x 10in. Dial 12 x 4in. All parts new and fully guaranteed. Complete with full construction data and operating manual.

LASKY'S PRICE 25 GNS. POST FREE
Also available ready built and tested 25 gns.
H.P. Terms available.

NEW MODEL SR-150



Covers full medium waveband and 1.6-4.4 Mc/s, 4.0-10.0 Mc/s and 11.0-3.0 Mc/s. In separate switched band spread ranges. Two aerials are fitted, an internal loop and external telescopic. Controls include: B.F.O. Sensitivity, C.W., A.N.L., tone switch, receive/stand-by, 8 meter. Easy to read illuminated dial with logging scale. For 200/250 V. A.C. 4 valves plus rectifier. Fitted with internal speaker and socket for phones or external speaker. Complete with full instruction manual. Cabinet size 12½ x 8½ x 6½ in.

LASKY'S PRICE 20 GNS. POST 10/-
H.P. Terms 25 dep. and 11 monthly payments at £11.14.0. Total H.P.P. £28.14.0.

MODEL HE80 12-valve super sensitive communication receiver. Freq. range 540 KHz-30 Mc/s. and 144-146 Mc/s. Dual conversion on 2 metres, with extra R.F. stage. Simple R.F. stage, two I.F. stages on all other bands. B.F.O. and Q-multiplier circuits. Improved A.N.L. and voltage regulated powerpack. "8" meter band spread on monster bands, large illuminated dial with logging scale. All controls fitted. Output for speaker and phones. Valve line-up: 4 x 6AQ5, 2 x 6BA6, 2 x 6BE6, 1 x 6BL6, 6AL5, 6AQ5, 6CA4 and 6AS. Steel case 17 x 7½ x 10in. For 200/250V. A.C. mains. Brand new with full instruction manual. No Kits available.

LASKY'S PRICE 59 GNS. POST FREE
H.P. Terms 25 dep. and 11 monthly payments at £4.18.0. Total H.P.P. £66.17.0.

STILL A FEW AVAILABLE Fully built and guaranteed MODEL BE30—LASKY'S PRICE 33 GNS. MODEL BE40—LASKY'S PRICE 18½ GNS. H.P. Terms available.

207 EDGWARE RD., LONDON, W.2
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Both open all day Saturday. Early closing Thursday.

Mail Orders and correspondence to 3-15 Cavell St., Tower Hamlets, London, E.1.



for multiplex adaptor. 5 valves—line up: ECC85, 6XN50L, ECC85, 6XN50L, ECC85, 6XN50L, ECC85, 6XN50L. Full vision tuning scale, size 21 x 6in. Overall dimensions 23 x 6½ x 8in. Made to the very highest standards.
LASKY'S PRICE 29½ GNS.



TRANSISTOR POCKET RADIOS
IDEAL GIFTS - All supplied complete with personal earpiece, battery and carrying case. Fully guaranteed—ready to use. **POST FREE.**
BOY'S 2 TRANSISTOR In attractive plastic case. Size only 4in. x 3¼in. x 1½in. Fitted 9½in. speaker. Socket for personal earpiece. Uses PP3 battery. Tunable over full medium waveband.
LASKY'S PRICE 35/-

BOY'S 4 TRANSISTOR MODEL
LASKY'S PRICE 45/-

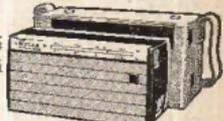
6 TRANSISTOR MODEL
Fully built in plastic case, 4in. x 9½in. x 1½in. with 2½in. speaker. Uses single PP3 type battery. Tunable over full medium waveband.
LASKY'S PRICE 59/6

8 Waveband (Long and Medium) Model. Size 8 x 3½ x 1½in. Cream/black plastic case. Including earpiece and carrying case.
LASKY'S PRICE 89/6

THE "TRANSWAVE" DE LUXE TWO WAVEBAND RADIO
8 Sony transistors give super sensitive reception over the full Med. and Long wavebands. Internal ferrite and detachable chrome telescopic aerials. 2½" diam. speaker. Smart black plastic cabinet with metal trim—size of 8 x 5 x 1½in. Uses 3 pen light batteries. Complete with leather case and shoulder strap, earpiece and batteries. Post 2/6.
LASKY'S PRICE 95/-

2 WONDERFUL SETS FROM THE U.S.S.R.

THE "SELGA" 7 TRANSISTOR RADIO
High sensitivity circuit gives excellent reception over full Med. waveband and powerful volume. P.M. speaker. Uses PP3 type battery. Black and ivory plastic cabinet, size 7 x 4 x 2in. Socket for external aerial. Complete with solid hide case and shoulder strap, earpiece and battery.
LASKY'S PRICE £5.9.6 Post 2/6



THE "SOKOL" 7 TRANSISTOR TWO BAND RADIO
Super quality receiver covering the full Med. and Long wavebands with exclusive rechargeable battery which can be recharged from any 120/250V. A.C. mains supply. Also operates on PP3 type dry battery. P.M. 8, outer Ivory and black plastic cabinet 8 x 3½ x 1½in. Socket for external aerial. Complete with real leather case, rechargeable battery and charger, dry battery and earpiece.
LASKY'S PRICE £6.9.6 Post 2/6

SPECIAL OFFER — MAKER'S SURPLUS 7 TRANSISTOR PORTABLE RECEIVERS

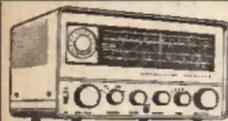


By famous British maker 7 transistor plus 3 diode superb, 8 waveband portable receiver. Operating from four 1.5 v. torch batteries. Cover the full Medium Waveband and Short Waveband 31-84 Mc. and also a separate switched band spread range, 10M, 10E, 10M and 20M with Band Spread Tuning. All Mutual Transistors and Diode Uses 4-UR batteries, 5in. Ceramic Magnet P.M. Speaker. Easy to read Dial scale, 100 MW Output, Telescopic Aerial & Ferrite Rod Aerials. Very attractively styled plastic cabinet size 10 x 6½ x 3½in. Finished in mid-lava with metal trim and carrying handle.

LASKY'S PRICE £10.19.6 Post 0/-
Four UR batteries 3/4 extra.

Tel: PAD 3271 | 42 TOTTENHAM CT. RD., LONDON, W.1 | Tel: LAN 2573
Tel: MUS 2605 | 152/3 FLEET STREET, LONDON, E.C.4 | Tel: FLE 2833
Both open all day Thursday. Early closing Saturday.

LASKY'S FOR SPEEDY MAIL ORDER SERVICE

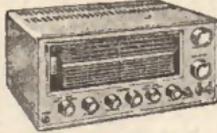


LAFAYETTE HA 43 COMMUNICATION RECEIVER

7 valves + Rectifier. 4 Bands 550 Kc/s - 1 Mc/s. "S" Meter. BFO-ANL-Bandspread Tuning. 200/250v. A.C. Brand New. 24 Gns. carr. paid.

HA.230 DE LUXE GENERAL COVERAGE RECEIVER

Wonderful value. 8 valves + rectifier. Coverage 550 kc/s - 30 Mc/s 1 R.F. and 2 IF stages. "Q" multiplier. B.F.O., A.N.L., "S" meter, electrical bandspread, aerial trimmer, etc. Brand new and guaranteed. £39. S.A.F. for full details. Also available in easy to assemble semi-kit form at 25 gns.



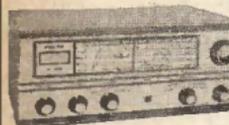
LAFAYETTE HA-55 AIRCRAFT RECEIVER

108-128 Mc/s. High selectivity and sensitivity. Incorporates 2 RF stages including 9CW4 Navigator. 8 tubes for 1 tube performance. Solid state power supply, adjustable squelch control, slide rule dial, built-in 4-in. speaker and front panel phone jack. 220/250v. A.C. Supplied brand new and guaranteed. £18.7.6. Carr. 10/-.



CODAR EQUIPMENT

Main London Agents
CR.70A. Communication Receiver £19.10.0
P.R.30. Freezerless. £5.10.0
A.S.6 Amateur Transmitter. £16.10.0
Postage extra.



TRANSORISED TWO-WAY TELEPHONE INTERCOM

Operative over amazingly long distances. Separate call and press-to-talk buttons. 2-wire connection. 1000's of applications. Beautifully finished in ebony. Supplied complete with basic & wall bracket. £8-10.0 pair, P.F. 9/6.

MAGNAVOX 163 Tape Decks
New 3 speed tape deck, supersedes old Coliaro studio deck.
2 track. £10-10.0, 4 track. £13-10.0

GARRARD AUTOCHANGERS

Autoslim £5.5.0
1000 £5.19.0
2000 £5.19.0
3000 £5.19.0
P. & P. 5/- extra.

AMERICAN TAPE

First grade quality American tapes. Brand new and guaranteed. Discounts for quantities.

3in. 225ft. L.P.	4/-
3in. 500ft. T.P. mylar	10/-
3in. 1000ft. std. plastic	8/6
3in. 500ft. L.P. acetate	10/-
3in. 1200ft. D.P. mylar	15/-
3in. 1000ft. T.P. mylar	55/-
3in. 1200ft. L.P. acetate	12/6
3in. 1800ft. D.P. mylar	22/6
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20mA	..	22/6	50V. DC	..	22/6
50mA	..	22/6	100V. DC	..	22/6
100mA	..	22/6	150V. DC	..	22/6
150mA	..	22/6	300V. DC	..	22/6
300mA	..	22/6	500V. DC	..	22/6
500mA	..	22/6	750V. DC	..	22/6
750mA	..	22/6	150V. AC	..	22/6
1-0-1mA	..	22/6	150V. AC	..	22/6
1-0-1mA	..	22/6	300V. AC	..	22/6
5A. DC	..	22/6	500V. AC	..	22/6
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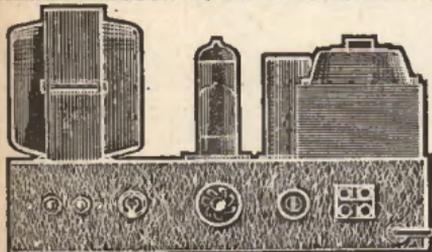
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FIVE-TEN AMPLIFIER

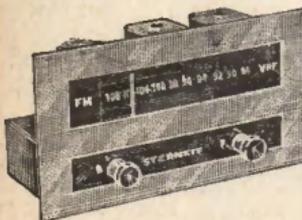
One of the most famous amplifiers ever made—thousands are in service as the nucleus of high quality installations throughout the world. Check the phenomenal performance, unattainable by any other amplifier obtainable in this price range: Freq. response 0 c/s. to 30 kHz. Sat. 1 db. Distortion 0.1% at full 10 watts output. Noise—68 db. below 10 watts. Input signal 40 mV. for 10 watts. Out. Imp. 8 or 16 ohms. Valves E251, RP96, EC038, 2 x EL84. Size overall: 10 x 7 x 4 1/2 in. high. Incorporates mains outlet for gram and sax. supply of 250 v. at 45 mA. and 4.5 v. at 2 1/2 amps. For pre-amp. tuner, etc. To full Mullard specifications. Assembled and tested £19.10.0 Kit £16. 0/6. **FIVE-TEN AMPLIFIER AND TWO-VALVE PRE-AMP.** Assembled and tested £21.10.0. Kit 18 0/6. **FIVE-TEN AMPLIFIER AND THREE-VALVE PRE-AMPLIFIER.** Assembled and tested £25.10.0. Kit 21.10.0. *Add 5/6 carriage.*

Especially low-priced, but actually a robustly made reliable instrument which can be used anywhere for note taking, idea recording, dictation, etc. Mechanism solidly built on to rigid steel chassis and housed in tough grey plastic case with carrying strap and clear plastic tape-deck cover. Picked up with single multiselector function switch and volume control. Hand microphone has stop/start thumbswitch for remote control of record and playback transcription. Complete with tape and spare spool, teflon and miniature earphone. Size only: 7 1/2 x 7 1/2 x 4 1/2 in. high. **£9.10.0. Carriage 4/6.**

NEW PORTABLE TAPE RECORDER



SUPERB NEW STERN-CLYNE FMI V.H.F. TUNER



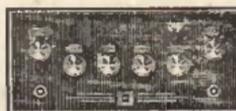
Especially designed for the Amateur builder, a new sensitive tuner unit that provides stable, interference-free reception of BBC FM transmissions. High quality output signal ensures optimum performance from any Hi-Fi audio system; superb styling makes for harmonious integration with existing equipment. Reliable easily aligned circuit employs BF

stage, tuned-anode tuned-grid freq. changer, 2 IP's, noise limiter and delay detector. Valves are 4 x RP91 plus 2 diodes. Input acc. 100mV for each. Dist. less than 1% at full deviation. Power reqd. 200v. at 50mA and 6.5v. at 1.5A. Panel black and silver-grey, size 8 x 5 1/2 in. Chassis: radiumina plated, overall depth 4 1/2 in. **FMI Kit of parts with Instruction Handbook, £7.8.0. Carriage 4/6. FMI Assembled and tested, £10.0.0. Carriage 4/6. Optional Power Pack Type D, Kit of parts, £21.5.0. Carriage 3/6. Power Pack Type D, Assembled and tested, £23.0.0. Carriage 3/6. Handbook only, 8/- post free. Descriptive leaflet on request.**



TWO-VALVE PRE-AMPLIFIER

Specially designed for use with the Mullard series of 3, 10 and 30 watt amplifiers but entirely suitable for all Hi-Fi power units but requiring input of more than 250 mV for full output. Features include inputs for crystal and variable reluctance pick-ups with RIAA equalization, plus radio, sensitive microphone, and tape replay direct or from pre-amp. Controls include 6-position selector, volume and wide range bass and treble boost. Valves 2 x RP96, size 9 1/2 x 4 1/2 x 3 1/2 in. high. Panel 10 1/2 x 9 1/2 in. Requires 300 v. at 3 mA. and 6.3 v. at 0.5 amp. Assembled and tested **£9.10.0. Kit 8.0.0. Add 5/- carriage.**



THREE-VALVE PRE-AMPLIFIER

Specially recommended for use with the 3-10 amplifier, but also suitable for use with all Mullard series music amplifiers and any high quality unit requiring an input of up to 250 mV. for full output. Features include: wide range Bass and Treble controls, high and low pass filters, auxiliary input and record output socket, switched inputs for crystal or magnetic pick-ups, radio, tape playback and auxiliary. Power required 250 v. at 6 mA. and 6.3 v. at 1 amp. Valves 2 x RP96 and EC038. Size 11 x 4 1/2 x 4 1/2 in. high. Panel 12 x 4 1/2 in. Assembled and tested **£13.10.0. Kit 11.10.0. Add 5/- carriage.**

TEST 7 Pocket Multimeter 39/6



Carriage 0/-

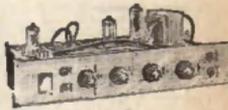
A really versatile instrument that makes a handy, pocket size tool. Measures AC or DC voltage in three ranges of 10-500-1000 v. Resistance 0-100,000 ohms, and Current 0-100 mA. D.C. Size only: 2 1/2 x 3 1/2 x 1 1/2 in. with ingenious dial design, providing a clear, easily read scale. Complete with battery and test leads.



MAGNAVOX 363 TAPE TRANSPORT

Manufactured to precise limits that permit recording on 1/4 inch tape back to the highest standards set by the Music Industry. Simple reliable design employs a single high-quality motor with heavy flywheel. Features include fast wind on and rapid rewind, pause control, 3-speed selection with interlock, built-in revolution indicator, piano key controls. Speeds 1 1/2, 3 1/2 and 7 1/2 L.P.S. Wow and flutter 0.1% on 7 1/2 L.P.S. Max. speed size 7 1/2. Playing time up to 130 min. per track from 1,200ft. standard tape. Size 13 1/2 x 1 1/2 in. plus 2 1/2 in. below mounting board. With 3 track heads. **£14.0.0. With 3 track heads. £21.0.0. Add 10/- carriage and insurance.**

HF/TR3 TAPE AMPLIFIER



Really the best complete tape amplifier available to the home builder. Supplied already matched for the Magnavox 363 tape deck.

Features include: switched equalization for all speeds (CNR standards at 7 1/2 L.P.S.); Treble boost incorporated during Record. Bass boost during playback, speaker output matched for 4, 7.5, and 16 ohms, additional outputs for extension speaker, phono monitoring on Record and Hi-Fi playback through existing systems, inputs for Mic. Pick-up, and VHF Radio; Valves: RP96, EC038, EL84, 6BM1, 6BM1. Size overall: 11 x 6 x 4 in. (Panel 13 1/2 x 4 in.). Power pack on separate chassis size 7 1/2 x 8 x 4 1/2 in. Amp. & power pack Kit of parts **£18.10.0. Assembled and tested £19. Add 7/6 carriage.**

TAPE PRE-AMP TYPE C



Specially developed by Mullard Laboratories for use with high quality replay systems, and supplied specially matched for use with the Magnavox 363 tape deck. Features include: ferronucleic pole core inductors for treble equalization, push pull oscillator incorporating ferronucleic transformer, adjustable output, for matching to existing high-quality amplifier systems, inputs for Mic. Pick-up, Radio, etc. Valves: 2 x RP96, EC038 and 6BM1. Totally enclosed in case size 11 1/2 x 4 1/2 x 4 1/2 in. (Panel 13 1/2 x 4 1/2 in.) Power supply of 300 v. at 20 mA. and 6.3 v. at 1.5 A. in separate chassis size 6 1/2 x 4 1/2 x 4 1/2 in. high to 2 1/2 in. below mounting board. Pre-amp and power pack Kit of parts **£18. Assembled and tested £19.10.0. Add 7/6 carriage.**

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*Type 2001—90 ranges, 50,000 ohms per volt sensitivity. Size only 2 1/2" x 4 1/2" x 1" deep with ingenious design allowing an extra large, easily read scale. Magnificently damped jewelled meter movement. Fitted with low loss, long life ceramic selector switch, complete with leads and test leads, battery and full instructions.

Ranges: DCV: 0.5-250-500-5000. ACV: 0-10-50-100-500-1000. DC Current: 0-500 uA, 0-5, 0.250mA. Resistance 0-5K-5M. Cap: 10pF-500pF and 0.001 uF-0.1 uF. Output: —30 to +10V.

WAS £5.50 NOW! £3.10.00 SAVE £1.15.00

*Add 2/6 p. & p.

BULK TAPE ERASERS



These are actually manufactured, high quality versions of the powerful demagnetizers used in instrument and aircraft manufacture, and represent the most effective way of restoring recorded or noisy tape to its original unmarked condition. Essential when using high fidelity recorders in conjunction with high-grade tapes: pays for itself over and over in preserving expensive tape performance. Bulk erasers can also be used to demagnetize small parts and tools, and for demonstration purposes, etc. Recommended types available:

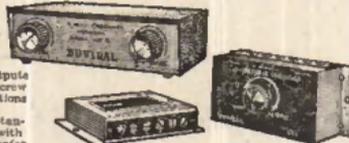
Model RT-120
Cleans up to 7in. reels of tape in seconds. For 250/250 v A.C. operation. Size: 4 1/2 x 4 x 9 1/2 in. high. Finished grey crackle.
WAS £5.5.00 NOW 6/16 SAVE £1.17.6

Model 178
Extra powerful professional model. Cleans tape reels in all sizes up to 10 1/2 in. diam. Size: 7 1/2 x 4 1/2 x 9 1/2 in. high. Finished grey crackle.
WAS £7.7.00 NOW £5 SAVE £2.7.00

1/2 p. & p. and Corv.

DIVIDAL CROSSOVER NETWORKS

Superb construction using selected high-grade components, each hand-soldered finished and constructed in metal cases with integral mounting feet. Inputs and speaker outputs are made to standard screw terminals. Full instructions supplied.



CV-1 Variable Crossover—Standard 3-way unit for use with low to medium power woofer and tweeter arrangement. Variable control permits adjustment of bass and treble output levels. In bronze finished case with black plastic panel size 4 1/2 x 3 1/2 x 1 1/2 in. high.

WAS £12.6 NOW 12/6 SAVE 10/- 3/4 p. & p. and Corv.

CV-2 Crossover Network—Hi-Fi 2-way iron-cored choke and filter system offering high frequency crossover at 2000 or 3500 cycles/sec. Totally enclosed in bronze finished case size 3 1/2 x 3 1/2 x 1 1/2 in. high.

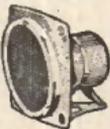
WAS £2.9.6 NOW 25/- SAVE £1.4.6 3/4 p. & p. and Corv.

CV-3 Variable Crossover—Standard 3-way, medium and high power variable level unit offering mid-range crossover at 2000 c/s and tweeter crossover at 3500 c/s. Together with "L" Pad variable level controls for mid-range and tweeter outputs. Matching arrangements also permits use as 2-way crossover, or without level controls when speaker efficiencies are equal. Hand-soldered finished in 9-tone grey metal case size 5 x 3 x 2 1/2 in. high.

WAS £4.10.00 NOW 50/- SAVE £2 3/8 p. & p. and Corv.

WESTWELL CT-2B TWEETER

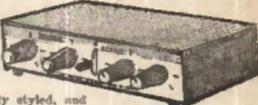
Adds sparkle and realism to the performance of almost any high quality reproducer. Nominal 1 1/2 diam. moving coil unit is attached to translucent bronze flange with matt black grille for mounting in or outside cabinet; alternatively detachable bracket can be used. Impedance 16 ohms. Rating 16 w max. 5 w nom. Recommended crossover frequency 3000 c/s. Sensitivity 105 dB/W. Size overall: 3 x 3 x 1 1/2 in. deep.



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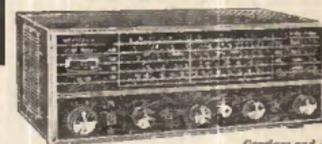
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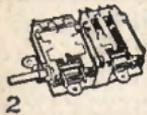
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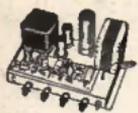
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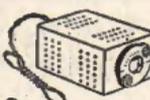
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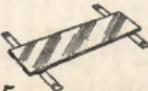
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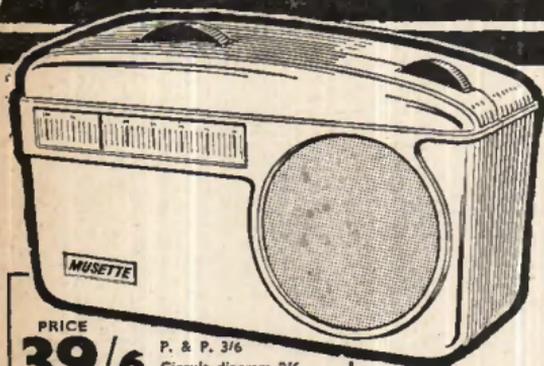
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PRACTICAL WIRELESS

APRIL 1966
VOL 41 NO 710

LAW OF THE JUNGLE

SOME readers take us to task for disapproving of pirate radio stations. A common accusation is that we are "against pop music"; that, teetering on the brink of senility, we cannot appreciate the prodigious talents of the Top Tenners. Such thrusts are wild indeed, for they completely miss the point!

Whether they churn out pop music, symphony concerts, language tuition or talks on advice to the lovelorn is beside the point; they should not be on the air at all.

Aircraft are not allowed to land and take off just where they fancy, flights must be controlled and scheduled—even though some of our critics claim that "the air is free for all to use how they like". Cars are not allowed to be driven at 100 mph in the wrong direction down a one-way street. This complex modern society insists that "liberty" must often be subjected to judicious control. And broadcasting is no exception.

Years of co-ordination and planning have made the jig-saw of the m.w. band fit together reasonably well. Frequencies, sites, power allocations have been arranged to the best mutual advantage. But the pirates burst in with a lofty disdain for law, order and the rights of other people to listen. Consequently protests have come from Sweden, Yugoslavia, Czechoslovakia, etc., concerning reception ruined by pirate stations.

Pirates also evade the Copyright Act, to the financial detriment of authors, composers and musicians. There is no control of programme or advertisement content. There is nothing to stop undesirable elements setting up their own pirate station. Unless . . .

The Government has, by implication, shown that the position of the pirates is indefensible. They are breaking international radio law. They are specifically contravening legislation signed last year by the Council of Europe—which other signatories have already used to shut down their pirates. Then what is the PMG waiting for—could it be a General Election!

Let him no longer tolerate the law of the jungle in broadcasting.

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Club Spot—Y.M.C.A. Belfast	1089

A New P.W.?

I READ PRACTICAL WIRELESS with great interest and would like to see you enlarge it. I would also like to see a PRACTICAL WIRELESS Annual (similar to the *Practical Motorist Annual*).

One other thing I would like to see is the revision of the *Practical Wireless Circuits* book, as I think this is very out of date—the coils and components being almost impossible to obtain.

I would like to see a licence for Novice Amateurs on 160m using 10W, as I think it would make use of a band which is dead most of the time.

J. Savage.

London,
S.E.11.

★ ★ ★

[PRACTICAL WIRELESS will be "going up in size" from the May, 1966 issue onwards, and *Practical Wireless Circuits* is, at the moment being revised and will contain up-to-date circuits using up-to-date components. It will be on sale in a few months time.]—Editor.

Correspondents' Club?

I HAVE noticed that there are several enthusiasts about my own age (14) who have written to P.W. asking for correspondents. I am sure that we could all get together and have regular correspondence, writing say, each month or fortnight to each in turn. By doing this, we would gain the experience of others and also offer our own experience. Some, like me, no doubt are preparing for the R.A.E. and would certainly benefit from corresponding with other enthusiasts.

Another idea I had was to swap components for stamps or other articles with enthusiasts in India, Ceylon, etc., who find it difficult to obtain electronic components and issues of P.W.

If anyone is interested would they please write to me. I will then write to all the members giving the names and addresses of the other members to start the "Pen Club" off. I would welcome any other ideas on the subject.

S. V. Odgear.

111 The Hollow,
Corsley,
Warminster,
Wiltshire.

NEWS AND..

P.W. AND P.T.V. FILMSHOW



Friday, 4th February, was the date of the P.W. Filmshow. Once again this year, the meeting was well attended by readers from all parts of the country.

In the absence of Mr. Stevens, the Editor, Mr. A. T. Collins, the Managing Editor of the Practical Group took the chair. After his opening speech, Mr. Collins introduced Mr. Ian Nicholson, of Mullard Ltd., the speaker of the evening.

Mr. Nicholson then introduced the first film, entitled "Electromagnetic Waves—Part 2". After this film there was an interval of 25 minutes during which refreshments were served.

During the second half of the programme, Mr. Nicholson gave a talk on Servicing Transistor Receivers. After this, there followed a film entitled "Thin-Film Microcircuits". The evening ended with a question and answer session during which Mr. Nicholson answered questions put to him by members of the audience.

The photograph shows Mr. A. T. Collins (left) and Mr. Ian Nicholson discussing a thin-film microcircuit.

AIRBORNE RADIO AIDS FOR EAST AFRICAN VC-10s
Three Super VC-10 aircraft on order for East African Airways are to be completely equipped with Marconi Sixty Series airborne radio navigation and communication aids, including the Doppler Navigator, which will form the primary en-route navigational system in the aircraft.

Each aircraft will be fitted with dual installations of the AD160 v.h.f. communications system, AD260 v.h.f. navigation system, AD360 automatic direction finder and the aircraft selective calling system, Selcal. Also in each aircraft will be the Marconi Doppler navigator, type AD560.

The AD560 Doppler navigator is currently in operation with BOAC, Qantas Airlines, Ghana Airways and Air New Zealand and will be fitted in new aircraft for Iraqi Airways and Pakistan International Airways in addition to the Anglo/French Concorde supersonic airliner.

.. COMMENT

ARCOLECTRIC RELEASE NEW CATALOGUE

Arcolectric Switches Ltd., Central Avenue, West Molesey, Surrey, announce their 1966 catalogue, No. 136. It describes their current range of switches, neon indicators, and signal lampholders.

Many new products are featured in this catalogue including the "27" and "28" range of lever and semi-rotary switches.

Copies of this catalogue are readily available upon request to Arcolectric Switches Ltd.

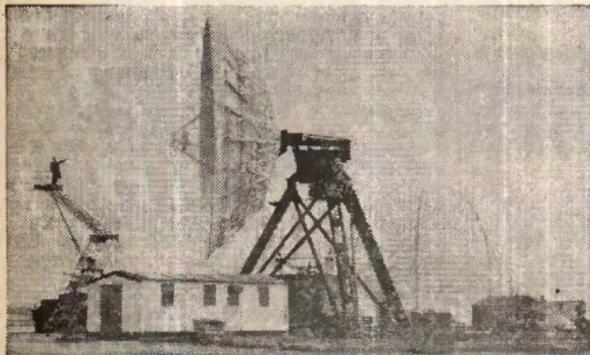
HARLOW MOBILE RALLY

The date of the Harlow and District Radio Society's Annual Mobile Rally will be Sunday, 25th September, 1966. Further details available from Hon. Sec., G. O'Donald, G3TLJ, "Great East", Harlow Road, Roydon, Harlow, Essex.

PETO SCOTT AMENDS NAME

Peto Scott Electrical Instruments Ltd. has changed its name to Peto Scott Ltd. It is thought that the former title gave a too restricted view of the wide range of activities now carried out by the company. The field of operation now extends through black and white and colour studio and professional television installations, professional tape recording, video tape recording, sound systems, functional music machines, Eidophor large screen television projection, language laboratories, teaching machines, schools TV equipment and overhead projectors.

SPACE ORDER FOR MARCONI



Marconi Co. Ltd., is currently building three space communications stations which are to provide the first British military satellite communications system. These stations are to take part in a joint project with the American military authorities, using a series of near-synchronous satellites launched by America.

This photograph, taken at one of the Marconi Company's experimental sites near Chelmsford, shows the 40ft. diameter dish aerial for the second station. It is seen here, fitted on a temporary mounting for test purposes. The white dome in the background is the 60ft. high inflatable radome into which the aerial will shortly be moved. The radome will provide full weather protection for the aerial, its fully steerable mounting, and for some of the associated equipment.

more News and Comment

The Meaning of Amateur

I HOPE the attitude of Mr. Davidson, G3FG (page 939, March issue) is not typical of the majority of radio enthusiasts.

I fully endorse the remarks of your Birmingham correspondent who G3FG criticises and would suggest that G3FG is being narrow-minded. It is, I think, possible to distinguish two separate hobbies (closely connected, I admit) i.e. set construction and DX (be it ham or broadcast DX). I myself could not really care who or what constructed my set as long as it gives me some DX. I know that some friends of mine will willingly construct a set and "home brew" it but once they get it working and calibrated, give no thought to the matter, take the set to pieces and start on a different circuit.

As to the meaning of "amateur" I hardly think it is relevant to either hobby. The thing that disturbs me most, from what I read of your correspondence, is the unwillingness of some—not many—to accept newcomers, novices not versed in the habit of set construction.

P. Charlton.

Middlesbrough,
Yorkshire.

Single Circuit Panels

I WAS intrigued by W. Groome's approach to the problem of assembling printed circuits without messy chemicals. (P.W. Feb., 1966). I wonder why he assumes that copper cladding has to be used?

I, as an "old timer" faced by this modern chassisless age—have found a simpler approach! I use 22 s.w.g. copper wire! What's more, on top of the board my assemblies look identical. Just as small and just as neat.

The only special tool needed is wire-bending pliers, to put neat loops on the ends of each link. The wires can run point-to-point by judicious use of bits of sleeving (though they needn't).

It's a lot easier to perform mods and de-bug new designs!

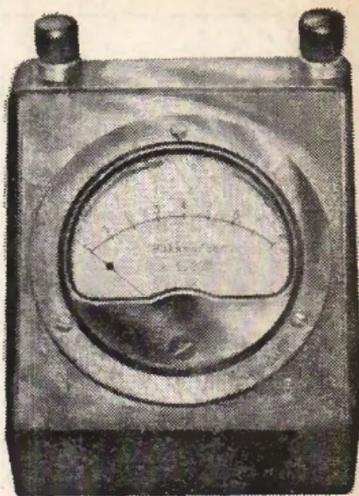
When finished, a layer of lacquer holds all in place . . . (Most times, surplus flux does all that is needed).

R. G. Young.

Peacehaven,
Sussex.

on page 1060

A Direct Reading Frequency Meter



★-Described by K. Royal

THE device about to be described commenced life in the author's laboratory for the measurement of the pulse repetition frequency of an oscilloscope's timebase in terms of a direct reading analogue on a milliammeter. This simply means that the milliammeter is scaled directly in terms of frequency instead of current. The meter circuit thus has to translate frequency to a direct-current value. Moreover, to be of much use it has to do this in a linear manner without progressive compression or expansion of the scale. That is, any change in applied frequency should give a correspondingly linear change in current reading over a specific frequency range.

The device then graduated to a "tachometer" for checking the speed of car engines, and finally, to a meter for indicating the frequency of any audio signal, whether of square, pulse or sine wave.

Frequency Discriminator

The key to the whole thing is a rather special frequency discriminator* whose circuit is given in Fig. 1. This is how it works. When the input signal goes negative, capacitor C1 charges through the signal source resistance, through R1 and through D1. With D1 connected the way round shown on the diagram is, of course, in forward conduction on a negative-going signal, while the emitter junction of the transistor Tr1 is in reverse conduction.

As long as the time constant $C1/R1$ is short (also assuming a low impedance or resistance source), C1 stores an electric charge equal to CV , where C is the value of the capacitor and V is the amplitude of the input signal.

Now, when the input signal changes to positive (or zero), C1 discharges through the emitter/base junction of the transistor and a pulse of current flows in the collector circuit. This happens on each cycle of signal, and since C1 discharge current is equal to $-CV/t$, where t is a time function of the

cycle and is itself equal to $1/f$, where f is the repetition frequency, it follows that the average emitter current is CVf . The average collector current is thus $CV\alpha$, where α is the current gain.

In the common base mode, α is almost equal to unity (actually it is less than unity because the emitter current is equal to the collector current plus the very small base current) so the average collector current can be considered to be CVf . From this, then, it can be seen that the current is proportional both to frequency and the value of C1. The frequency range over which the discriminator will function is thus influenced by the value of C1. Table I relates maximum frequency to value of C1.

In Fig. 1 a milliammeter is included in the collector circuit, and this is scaled to read frequency direct. It is possible to employ a high resistance voltmeter instead of a current meter by putting a resistor in series with the collector and using the voltmeter to indicate the voltage in terms of frequency developed across it. In this case, the voltage as recorded would be equal to $CVfR$, where R is the value of the collector resistor.

Input Limiter

The basic circuit in Fig. 1 is dependent upon the signal source impedance and amplitude, and is thus somewhat unpredictable. For more consistent operation a "buffer" stage should be interposed between the signal source and the discriminator,

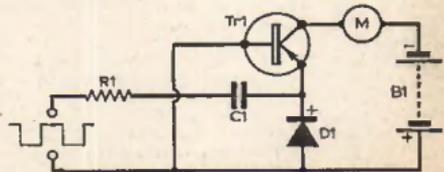


Fig. 1: Circuit of the basic frequency discriminator.

with the buffer acting also as an amplitude limiter.

A practical arrangement in this context is given in the circuit in Fig. 2. Here Tr1 is the buffer/amplitude limiter. It serves virtually as a switching transistor the input signal performing the switching action. This is how it works.

On a negative-going input signal current flows in the emitter/base junction of Tr1 causing a flow of current in the collector resistor R3 and the voltage at the collector to fall. On a positive going input signal, the base is made positive with respect to the emitter and no current flows in the collector resistor and the voltage at the collector rises almost to the supply voltage. During this half-cycle, however, D1 conducts and maintains a constant mark/space ratio. That is, the transistor is switched on and off for equal periods of time.

A square-wave is thus developed at the collector of Tr1, and since the amplitude of this wave is held constant by the transistor bottoming when the input signal swings negative and cutting off when it swings positive, amplitude limiting occurs provided the input signal amplitude is itself sufficient fully to switch the transistor as described. Thus, any increase in signal input amplitude does not influence the amplitude of the square-wave signal at the collector.

The signal, then, is of ideal form to work the frequency discriminator which, in Fig. 2, is Tr2, with C2 as the charging capacitor. A preset resistor is included in series with Tr2 collector and milliammeter to set the current for full-scale deflection at the top frequency it is required to indicate.

Unit Construction

This circuit, in fact, was used by the author both to indicate the pulse repetition frequency of the timebase of an oscilloscope and to determine the turnover speed of a car engine by "counting" the number of pulses produced by the contact breaker of the ignition system, the scale of the milliammeter then translating these direct to revs per minute.

The circuit was built upon a small piece of "Eyelet Board" measuring about $2\frac{1}{2} \times 2\frac{1}{2}$ in. The special "eyelets" (see the Component List) are secured in the holes corresponding to the component lead-out wires. The wires, along with the circuit connecting wires themselves, are then soldered to the eyelets.

A word or two about the eyelet board would not be amiss at this juncture. The board itself is made of a resin-bonded laminate and the eyelets are of tinned brass. The laminate is perforated every 0.2 in. and the holes (which are about $\frac{1}{16}$ in. in diameter) tightly accommodate the eyelets. These are prevented from being pushed right through the holes by a slightly raised flange at one end. This also gives a little clearance for winding round the circuit connecting wire.

The eyelets are fitted to the selected holes by inserting the barrel end to the flange, turning the board over and then gently tapping the protruding barrel end with the tip of a centre punch. One of the automatic press-type punches allows this operation to be handled swiftly and accurately.

Meter Connection

Fig. 3 shows how the board is processed to accommodate all of the components of the circuit. These are wired together underneath the board and connection to the meter terminals is accomplished by the positioning of eyelets close to two holes made in the board to take the terminals of the meter. Each hole thus has one meter connecting eyelet close to it, as shown in Fig. 3. Washers placed on the meter terminals after they have been pushed through the holes in the board thus provide good electrical connection between the adjacent

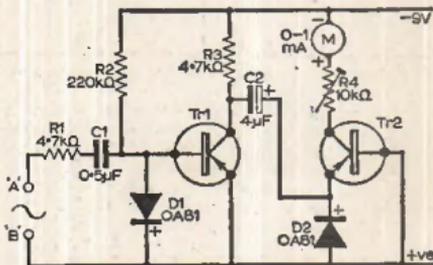
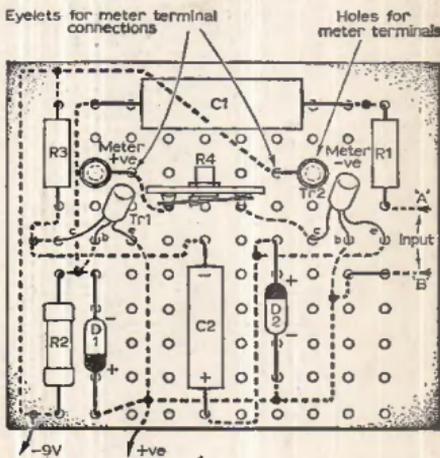


Fig. 2 (above): Circuit of the instrument described in the text.

Fig. 3 (below): Layout of components on the eyelet board.



eyelets and the terminals when the terminal nuts are tightly screwed down. In this way, therefore, the module is secured to the meter movement.

The 1mA meter movement employed by the author had its own mounting case and this easily accommodated the meter-mounted module with room to spare for a small PP3 or PP4 battery if required. The author's prototype is externally powered so no on/off switch was fitted. However, a switch would be needed should internal battery powering be adopted, since the limiter takes a small quiescent current of just under 1mA.

The front view of the instrument, with the scale calibrated in pulses per second, is shown in the heading photograph. The two top terminals accept the input signal and supply positive, to which the input signal is relative while a small terminal on the rear of the case picks up supply negative.

TABLE I

Maximum Frequency	Value for C2
100 c/s	1 μ F
1 kc/s	0.1 μ F
10 kc/s	0.01 μ F
100 kc/s	0.001 μ F
1 Mc/s	0.0001 μ F

Table 1: Showing maximum frequency for variations of C2.

As mentioned earlier the instrument was made originally to indicate timebase repetition frequencies of an oscilloscope up to a maximum of 8kc/s. For this application, a value of 0.01 μ F was found suitable for C2 (see Table I). After setting the timebase to 8kc/s as determined by a calibrated audio oscillator, the preset R4 should be adjusted to give a deflection of 8 on the meter. A signal of any frequency below 8kc/s is then read directly from the scale.

As a Rev-counter

For use as a car engine rev-counter, the pulses as developed between the contact breaker (CB) and switch (SW) terminals of the ignition coil are applied to the instrument. The pulses here have a rate proportional to the engine speed. For a four-stroke engine, the frequency is equal to engine

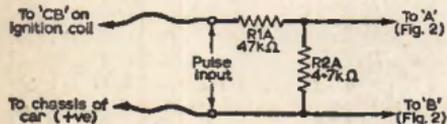
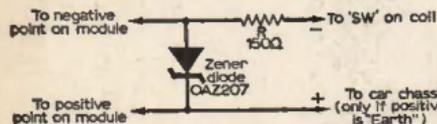


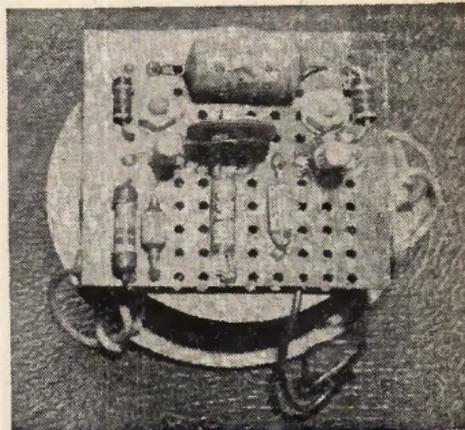
Fig. 4 (above): Circuit of input attenuator required for indicating ignition pulse rate.

Fig. 5 (below): Circuit of zener diode stabilisation.



turnover speed in revs per minute $\times n/120$, where n is the number of cylinders. Thus, such an engine running at, say, 3,000 r.p.m. produces a pulse rate of 100 c/s, while a single cylinder, four-stroke motor-cycle engine will produce a pulse rate of only 25c/s at the same turnover speed.

A top engine speed of 6,000 or 8,000 r.p.m. is generally conventional depending really on the nature of the engine. Thus, the value for C2 is usually in the order of 1 μ F, but for single cylinder engines a value up to 4 μ F may have to be used. The value in these cases is best determined by experiment for the least ripple effect on the needle at low revs.



Rear view of finished instrument showing completed circuit board.

It is not safe to apply the contact breaker pulses direct to the circuit since their large amplitude could damage the semiconductors. An attenuator after the style of that shown in Fig. 4 circuit should be used at the input. The rev-counter may be powered either by its own, internal battery (as may be required for some motor cycles without battery powering) in which case the "pulse input" point is connected to "CB" on the ignition coil and the supply positive line of the circuit is connected to the metal of the engine or car body, or powering can be from the car battery.

In the latter case, the car supply voltage must be stabilised, as this has a tendency to rise and fall with engine revs and with varying loads on the electrical system.

The best stabilising element for this application is the zener diode. When this kind of diode is biased for reverse conduction it becomes a relatively low resistance at a specific value of reverse voltage, called the "zener voltage". The current then passed by the diode is called the "zener current". Under this condition, the voltage developed across the zener diode remains substantially constant over a range of input voltages. This means, then, that the voltage across the diode is stabilised.

A zener diode stabilising circuit is given in Fig. 5 and "R" in series with the supply and the

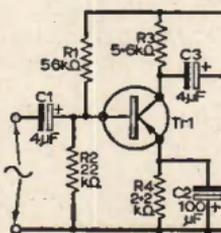


Fig. 6: Circuit of common-emitter amplifier for increasing the sensitivity of the instrument, as described in the text.

COMPONENTS LIST**CIRCUIT FIG. 2**Resistors (all $\frac{1}{2}$ -watt carbon insulated)

- R1 4.7k Ω
 R2 220k Ω
 R3 4.7k Ω
 R4 10k Ω preset (printed-circuit board type)

Capacitors

- C1 0.5 μ F 12v
 C2 (see Table 1)

Semiconductors

- Tr1 and Tr2, OC71
 D1 and D2, OA81

Meter Movement

Moving-coil 0-1 mA

Sundries

Eyelet Board and eyelets (available from Messrs. R. & E. Lamb, 17 Queens Road, Leytonstone, London E.11. Connecting wire, battery (PP4) and battery clips. Instrument case.

CIRCUIT FIG. 5**Resistor**150 Ω 1-watt**Semiconductor**

Zener diode, OAZ207

CIRCUIT FIG. 6Resistors (all $\frac{1}{2}$ -watt carbon, insulated)

- R1 56k Ω
 R2 22k Ω
 R3 5.6k Ω
 R4 2.2k Ω

Capacitors

- C1 and C3 4 μ F, 12v. Electrolytic
 C2 100 μ F, 6V. Electrolytic

Semiconductor

Tr1, OC71

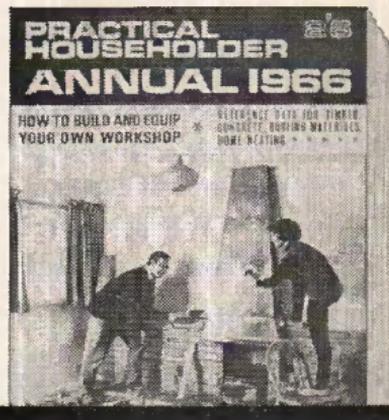
diode sets the zener current. The value is determined by the required load current, the supply voltage and the type of zener diode used. The manner of connecting to the car is revealed in Figs. 4 and 5. This applies only to cars with 12V supplies and positive earths.

The main application of the device, of course, is as a direct reading frequency meter, and the way that it can be used to indicate engine revs may be of academic interest only to some of our readers. Nevertheless, the versatility of the device is revealed.

To provide a frequency indication from low-level signals, an amplifier is required in front of the main circuit of Fig. 2. A suitable amplifier is given in Fig. 6. This is just an ordinary common-emitter circuit, but the lift that it gives to signals is sufficient to allow the device to respond to signals as low as 50-100 mV. When this amplifier is used, R1 and C1 from Fig. 2 circuit should be omitted. ■

* D. E. O'N. Waddington, "A Simple Frequency/Voltage Converter" Marconi Instrumentation, Vol. 10 No. 1.

JUST OUT!



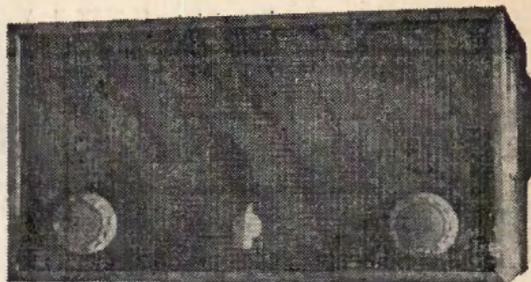
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2/6 NEWSAGENTS AND BOOKSTALLS

A
TUNED
SWITCH-
SECTIONS



SUPERHET

F. L. THURSTON

PART 2

STAGE 3—AERIAL AND PRE-SELECTOR SECTIONS

A 10 x $\frac{1}{4}$ in. ferrite rod is used in the aerial, which is home wound. While this is recommended, it is not essential, and a ready-made type of aerial may be used if preferred. The winding and mounting details for the home-made type are shown in Fig. 5. Note that the aerial is mounted directly on the speaker.

Secure the external aerial/a.g.c. socket in place at the rear of the chassis, checking that there are no shorts to chassis. Secure L3 in place above chassis.

Secure S2 in place.* It will probably be necessary to make this switch up by modifying an alternative type, and the essential dimensions are given in Fig. 6. Such a conversion is not difficult: a multibank switch, at least as long as that shown, is obtained. The switch must have at least 4-ways. The switch is then dismantled, reduced to the correct length, and re-assembled to conform to Fig. 6. If too many "ways" are available, they can be reduced to the correct number (4) by drilling and tapping a 6BA hole in the switch front to take a short 6BA screw which will act as a stop.

When the switch has been secured in place at the front of the chassis, secure the trimmer capacitors in place, as shown in Fig. 6. The method of mounting the beehive trimmers is shown in Fig. 7, where it can be seen that they are soldered to the drilled-out ends of 2 BA screws, which are then bolted directly to the chassis, using a solder tag in place of a normal washer.

Now wire-up the circuit as shown in Fig. 6. Use screened leads where indicated. Check the wiring and switch on. Test voltages are V1, anode 250V, screen 145V, cathode 1.5V.

*S2 can be made to order by: Specialist Switches Ltd., 79a Duke Road, London, W.4. Price approximately 17/6. Drawings must be supplied with order.

TESTING

Temporarily connect a germanium diode in circuit with one end to the loose end of C10 (audio input) and the other end to the grid of V1. Switch S2 to position 3. The ferrite aerial and diode are now connected as a simple crystal set. If the volume control is turned full up, the Light programme should be heard reasonably well. By adjusting C23 it should be possible to tune the programme in quite sharply, although the Home will probably be heard as well.

Now disconnect the diode from V1 grid and connect it to tag 2 of the 16-way tag strip. The circuit is now connected as a t.r.f. receiver. When S2 is switched to position 3 it will be seen that, by adjusting C23 and C26 together, the Light can be tuned in far more sharply than before, that interference from the Home is greatly reduced, and that the Light now comes in with vastly improved strength. By connecting an external aerial (a few yards of wire into the external aerial socket, the signal strength is increased even more.

The trimmer capacitors should now be aligned, in conjunction with positions of S2, as follows:

- Position 1—(Radio Caroline). Tune C21 and Slug of L3.
- " 2—(Luxembourg). Tune C24.
- " 3—(Light). Tune C23 and C26.
- " 4—(Home). Tune C22 and C27. Fixed padder.

Note that the above tuning sequence must be followed. The stations on positions 1 and 2 may be very weak.

This concludes stage 3.

STAGE 4—OSCILLATOR AND MIXER SECTIONS

Secure L4, the oscillator coil in place below the chassis. This coil is sold as a short-wave oscillator coil, to cover the 90 to 250 metre bands. At the particular frequencies used in this receiver, this coil gives a better L/C ratio, and therefore better frequency stability, than would a normal medium-wave oscillator coil.

with no leads soldered to it). Adjust the lower tuning slug of i.f.t.1 for maximum signal (any position of S2 may be used). Now re-trim upper tuning slug, and finally re-trim lower slug again. The primary of i.f.t.1 is now fully aligned, while the secondary may be considered to be temporarily aligned (it will require re-trimming after the two additional i.f. stages are added in stage 5 of the construction).

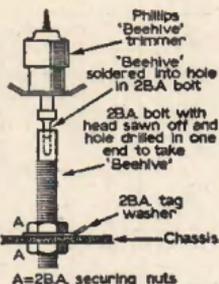


Fig. 7: Method of mounting the beehive trimmers.

EXPERIMENTS WITH THE MIXER/OSCILLATOR SECTION

As constructed, the circuit used for the mixer gives very low noise, while the oscillator is exceptionally drift free and immune from "pulling". By making temporary alterations to the circuits, it is possible to demonstrate quite forcibly the inferiority of some alternative circuits. Space is limited, however, and only two experiments will be given, as follows:

If a small transistor radio with built-in aerial is available, place it near to the oscillator coil (L4) and tune through the medium waveband; a number of whistles will be picked up, demonstrating that the oscillator is working. Tune in to one of these whistles, and then move the transistor set to the r.f. coil, L3; the whistle will still be there, but will be very weak. Finally, move the transistor set to the ferrite aerial; if any whistles are still available they will be very weak indeed.

Now unsolder the 1000pF capacitor, C18, from pin 2 of V2 (mixer cathode) shown as a temporary connection in Fig. 8. Solder a 15pF capacitor between pin 1 of V3 oscillator anode) and pin 1 of V2 (mixer grid). The circuit is now connected as a mixer with the oscillator injected to the grid, quite a common circuit.

If the transistor set is now placed near the r.f. coil, L3, whistles will be heard far more strongly than was formerly the case. Whistles will also be heard at the ferrite aerial. It may be noted that if a superhet is made without a pre-selector stage, using oscillator injection to the mixer grid as is often done the resulting "receiver" will in fact act as a quite effective transmitter!

Another point may be demonstrated with this circuit, by pressing a finger against the

mixer grid pin; a quite considerable reduction in volume will be noticed. This loss of volume can be attributed to loading and de-tuning of oscillator and r.f. stages. If the circuit is now re-wired to conform to Fig. 8 (the "temporary" connections can now be made "permanent") and the same finger test is made to the point of injection (mixer cathode), it will now be found that no reduction in volume takes place. This is because both the oscillator and r.f. stages are isolated from the point of injection, and also because no loading of the actual injection voltage takes place due to the low impedance used.

The mixer gives very low noise and good gain, and an essential part of the design which makes it possible to obtain these results is the voltage divider and decoupling network, R3-R4-C3, which feeds a constant and predetermined voltage to the mixer screen grid; the screen grid voltage must be within 5V of that shown in Fig. 1. An effective way of demonstrating the importance of this voltage is to replace R3 with a variable resistor of about 100 to 250kΩ value, and adjusting the value while monitoring the screen grid voltage; it will be found that as the voltage departs from that specified, the noise increases and the gain decreases. Many conventional mixers are found to have the high noise and low gain that is associated with this test. Once the circuit has been re-wired to conform to Fig. 8, stage 4 of the construction is complete.

STAGE 5—I.F. SECTIONS

Wire up the circuit as shown in Fig. 9. Check wiring. Switch on and check voltages as shown in Fig. 1.

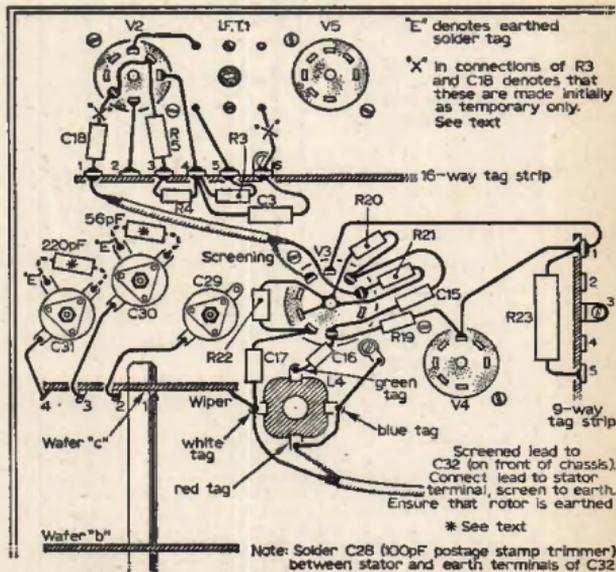


Fig. 8: Wiring and layout for stage 4 of construction.

NOTE. Points marked X in Fig. 9 are temporary connections only.

ALIGNMENT. Disconnect the capacitor used in aligning stage 4 from the secondary of i.f.t.1, and reconnect to the primary of i.f.t.2 (connected to pin 5 of V5). Adjust the primary tuning slug for maximum signal, then re-trim secondary of i.f.t.1 again, and finally re-tune i.f.t.2 primary again. Alignment of i.f.t.1 is now complete.

Move the capacitor to the secondary of i.f.t.2, and repeat the above procedure with the primary and secondary tuning slugs of i.f.t.2. When alignment is satisfactory, move the capacitor to the primary of i.f.t.3, and align the secondary slug of i.f.t.2 and the primary slug of i.f.t.3. To complete the tuning, move the capacitor to the secondary of i.f.t.3 and trim the primary and secondary windings of that i.f.t.

Note that, because of the use of three i.f. stages, the tuning is so sharp that it will probably be necessary to de-tune the i.f.'s slightly if acceptable quality audio is to be received.

Stage 5 of the construction is now complete.

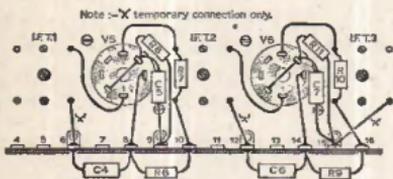


Fig. 9: Wiring and layout for stage 5 of construction

STAGE 6—DETECTOR AND A.G.C.

The permanent detector and filter network can now be connected into the circuit.

Remove the temporary detector circuit used in the two preceding stages (the capacitor and diode). Wire up the circuit as shown in Fig. 10.

The dimensions of the component board shown are approximate only; as an alternative to the board, the components may be soldered to suitable

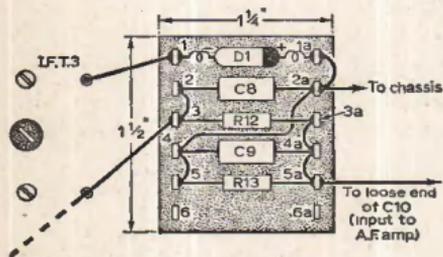


Fig. 10: Wiring and layout of detector/filter.

miniature tag strips. When completed, the circuit is switched on to test that it is working correctly.

If working satisfactorily, wire up the a.g.c. circuit as shown in Fig. 11. Note that the "tem-

porary" connections of Fig. 9 are removed and re-wired as "permanent ones as in Fig. 11. After checking over the wiring, switch on and select a station. The a.g.c. circuit may result in some small loss in volume compared to that obtained earlier; the stations should now, however, be virtually free from "fade".

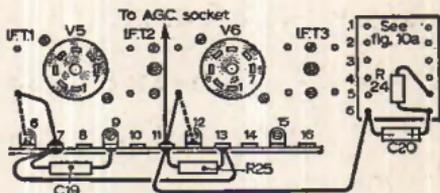


Fig. 11: A.G.C. wiring.

If a voltmeter is available it can be connected to the a.g.c. socket provided at the rear of the chassis. The a.g.c. voltage is negative, and rises in magnitude with the strength of the i.f. signal. Thus, as a signal fades, so the a.g.c. voltage will fall. The voltage will vary between about zero and 5V (measured with a v.t.v.m.), the normal amplitude being about 4V.

The tuning of the r.f. oscillator, and i.f. stages can be perfected if required, by using the a.g.c. point as a tuning monitor.

The construction of the receiver is now complete.

THE CABINET

The prototype receiver was placed in a simple, close fitting cabinet, which was suspended below a cupboard in the kitchen, as shown in the photograph. This cabinet may be a trifle small for really good, cool running of the radio, and the reader may prefer to make a similar cabinet but of larger dimensions. Quarter-inch or heavier ply-wood is used in the construction. The sides and top and base are cut to the required size and nailed and glued together. A fairly rough form of construction can be used, as any mistakes are "camouflaged" when the unit is finally covered with rexine.

Before proceeding further with this "box" section of the cabinet, make up the front panel and check that it fits correctly to the radio. When making the front panel, allow for the thickness of the speaker gauze that will be used to cover it, and the thickness of the Rexine used to cover the "box", when working out the overall dimensions. When the front panel is correctly cut, cover its front and sides with the selected speaker gauze, gluing it in place with "Copydex" adhesive.

Now place the Rexine selected to cover the "box" in position, and after trimming, glue (using Copydex) the Rexine to the front edge and inside only. When the glue has set (about 10 minutes) push the completed front panel in place in the box, peel back the Rexine layed on the box, and secure the front panel in place with a few light nails, hammered in from the outside of the box.

Finally, complete the cabinet by gluing the remaining Rexine in place. ■

on the Short Waves

MONTHLY NEWS FOR DX LISTENERS

All times are in G.M.T.

All frequencies are in kc/s.

The Broadcast Bands—by John Guttridge

CONTRIBUTIONS have been lighter than usual this month. Don't forget we want your news. Thanks go to J. McNally, S. Haagensen, S.B.C., A. J. Jenkins, Middlesbrough Boys High School S.W. Club, I. Black and P. A. Church.

Algeria: *Radio Algerie* (21 Boulevard des Martyrs, Algiers) now has English from 2200—2230 on 6,175/890. Same outlets are used for Spanish 2230—2300 and French 0630—0830 and 1700—2200. French is also carried from 1200—1700 (0900—1700 Sundays) or 890/11,835. Report forms are now available from this station which now gives full QSL verification.

Congo: *Radiodiffusion de la Republique Democratique du Congo*, (B.P.3171, Leopoldville) has been heard in French at 1900 on the new frequency of 9,780. Local dialect follows at 1927.

South Africa: *South African Broadcasting Corporation* (P.O. Box 8606, Johannesburg) now carries the Africa service at 0300—0400 on 6,150/7,270, 1000—1600, 15,220/17,805; 1600—1710, 4,975/15,220/17,805, 1710—1845, 4,975/11,900, 15,220; 1845—2000, 4,975/9,525/11,900; 2000—2100, 4,975/7,270/9,525; 2100—2115 7,270/9,525. On Sundays transmissions start at 1100 and 17,805 changes to 11,900 at 1650. One of the stations new 250kW outlets is used on 4,975. At 2000 11,900 carries the programme *Radio Paradys*.

Tunisia: *Radiodiffusion Television Tunisienne* (139 Avenue de Paris, Tunis) is now carrying the home service on the new frequency of 6,305.

Burma: *Burma Broadcasting Service* (Promo Road, Kamayut P.O., Rangoon) can be heard from around 1300 to sign off at 1345 on 4,795.

Japan: *N.H.K.* (Tokyo) can be heard with English around 1300 on 9,525, though there is interference from *Voice of America*, Greenville, in Spanish to Latin America on the same frequency.

Lebanon: *Lebanese Broadcasting Station* (Radio Lebanon, Ministry of Orientation, Information and Tourism, Beirut) has been reported back on 11,770 to Africa from 1830—2030 (English 1830—1900). Reception in England is good.

Papua and New Guinea: *Radio Rabaul* (P.O. Box 71, Rabaul, New Britain) uses 3,385 from 0600—1300. Sends acknowledgement folder. *Radio Kerema* (Gulf District, Papua) on 3,245 from 0720—1100. *Radio Daru* (Western District, Papua) on 3,304 from 0800—1100. *Radio Goroka* (Eastern Highlands, New Guinea) on 2,410 from 0800—1100.

U.S.A. *Voice of America* (US Information Agency, 330 Independence Avenue, S.W. Washington 25, DC) will use the following frequencies to Europe in English from March 6, 3,980 (Munich) 0300—0730, 1400—2345; 5,965 0300—0730, 1630—2200, 5,995 (Greenville) 0300—0730; 6,040 0500—0700; 7,200/7,250 0330—0730; 7,205 1500—1900; 7,210 1900—

2245; 9,540/9,740/15,295 0430—0730; 9,565 1830—2245; 9,760 1700—2245; 11,760 1830—2215; 15,205 (Greenville) 1400—2215; 15,290 1400—2000; 17,780 (Greenville) 1400—1800; 1,196 (Munich) 1600—1830.

Netherlands Antilles *Trans World Radio* (Bonaire) has English at 2100 in the 25m.b. Frequency is believed to be 11,840.

Argentina *Radiodifusora Argentina al Exterior* (RAE), (Sarmiento 151, Buenos Aires) now carries English from 2300—2345 and not 2400 on 11,710. The programme has also been occasionally heard on 11,780 and 9,690. After close down of the International service at 0030, 11,710 is reported to carry the National Service.

Brazil: The following stations are available in the early evening: *Radio Mayrink Veiga* (Rua Mayrink Veija 15, Rio de Janeiro) on 11,770; *Radio Sociedad de Bahia* (Rua Carlos Gomes 57, Sa Ivador, Bahia) on 11,875; *Radio Tupi* (Avenije Venezuela 43, Rio de Janeiro) on the new frequencies of 11,705/9,800; *Radio Bandeirantes* (C.P.372, Sao Paulo) on 11,925; *Radio Brasil Central* (C.P.330, Goiania) on 11,815.

Denmark: *Radio Denmark*, (Radio House, Copenhagen V) may reintroduce the programme "Short-waving to the World" on Saturdays at 1015—1100 on 9,520. On March 6, 15,165 replaces 9,520 for the 1730—1810 transmission.

German Federal Republic, *Deutsche Welle* (Bruedesstrasse 1, Postfach 344, 5 Köln) now uses 9,575/7,175 for the 0300—0340 English transmission.

German Democratic Republic, *Radio Berlin International* (Berlin-Oberschonweide, Nalepastrasse 18-50) uses the additional frequency of 5,300 in their English European transmission at 1730. Other frequencies are 6,080/6,115/7,185/7,300/9,730. There is a new transmission at 1815 on 1,511.

Great Britain, *Manx Radio* (P.O. Box 22, Douglas, Isle of Man) is understood now to have a power of 2kW. After dark uses 1,594.

Hungary: *Radio Budapest* (Brody Sandar—S.U. 5-7 Budapest VIII) has English to Europe on 5,900/7,220/7,305 from 1930—2000 and 2200—2230. The 2200 transmission is also carried on 6,234. English to North America is at 0030—0100, 0130—0230 on 9,833/9,540/7,220/6,234 and 0300—0400, 0430—0500 on 9,833/7,220/6,234/5,900.

Portugal: *Radio Lisbon*, (Rua Sao Marcal, 1-A, Lisbon) reported using 7,285/7,130/6,185/6,025/5,975 for English to Europe at 2015—2100.

Egypt: *Cairo Radio* (UAR Broadcasting and TV, Maspero, Cairo) now carries English to Europe at 2100—2215. Frequencies are 9,475/11,915.

Saudi Arabia: *Saudi Arabia Broadcasting* (Ministry of Information, Airport Road, Jeddah) puts in a strong signal after 2200 on 7,220 but suffers from Budapest on same frequency and a jammer on 7,215.

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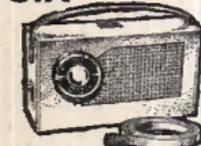
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MELODY SIX

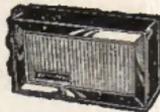


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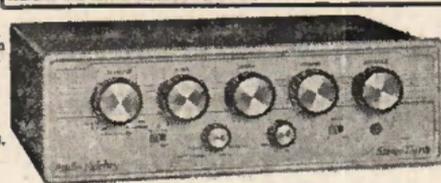
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WEIGHT 12lb.

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m.v.

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The Amateur Bands—by David Gibson G3JDG

WANTED! James Bond type agents to sniff out the truth. Out of a large mail bag this month most letters were about 20 DX? No, nor about the VK on topband, or the openings on 20 either. They referred to "that man" NS1A but with a new twist. Apparently we now have a NS1B and NS1C too. No, they do not count as separate zones for a countries worked certificate. Briefly the locations of these stations are assured as follows:—On board Radio London; 10 miles south of Herne Bay; in a motor boat with a No. 19 transceiver; on an Irish river (up the proverbial creek perchance). Will the real NS1s please stand up.

I am taken to task about 80 being deserted by George, ZL3QX, who informs that both ZL4IE and ZL3FZ have been consistently working EUs for some months. Two calls for 80 metre sleuths to be on the watch for. The l.f. bands are producing some very good openings at the time of writing. Both phone and c.w. are well represented—go on, live dangerously, listen now on 160, 80, and 40.

Forty and Down

Some very fine logs for 160 shows what's about. **D. Douglas** (Dundee), HRO, 60ft. 1-wire, sends in what might be confused for twenty, but don't be fooled, it is 160, G, GI, GM, GW, GD, DL1FF, DL4SS, EP21W, HK4EB, JA6AK, KV4CI, OE6HS, YV0AA, ZB2A, 9M6BM, R. Iball (Workop), SX28, PR30 pre-selector, 160 raised IS1F, K2DGT, VE2ITU, VE3-AGX, BWY, QU, VO1HN, W1-BB/1, HGT, W2GGL, W8ANO, W9YYG, W0VXO. **Steve Wilson** (Ossett), BC348R, 264ft. 1-wire, 160—DJ2GL, DL7FZ, GI, GM, GW, GD, HB9-CM, TT, IS1FR, OK1-AEF, AKO, OLS-ADO, AFF, OE1FLW/1, OE6HS, PA0DC, PA0PN, VE3BWW, VO1-FB, HN, W1-BB/1, JJU, W8HGW, W0VXO, ZB2AM, 9H1AE. On 40 Steve's best were AP2LK (Pakistan), CN8AW, HK7BCX, VP6BX. On 80 things are lively too, **J. Brown** (Llandaff), 19 Set, dipole, raised the following on phone, CN8AW, CT1-SQ, EE, IT1-AUL, ZGY, MP4BAA (c.w.), many VE's, VP-5GC, 7NP, 9AV, W's including W8-BON, WNL, YN4CU, ZB2AJ, 4X4BO, 7X2AH, 9J2MX. Again, on 40, **Steve Wilson** logged CM2WS, CO1PY, CR7CI (Mozambique), CX2CO (Uruguay), EP2BQ, FG7ND, HK-3AST, 3UZJ, 4ADD, 4ALE, HV1CN (Vatican), JA1OHV, K2-GL, KBT, PCB, K4JC, K0LZU, KG4AN, KP4BBN, LU6FA, PY-1, 2, 3, 5, 6, 7, 8, VP5AR (Turks and Caicos Is.), VP9FT, VP7NW, W3-MGK, CRQ, W4LUV, WB6JJC/M, YV4CJY, ZC4GB, 4X4FA, 5A3TX, 6Y5XG, 9G1FQ, all c.w. **N. Flatman** (Ipswich) R3629, 68ft 1-wire, 80 metres bore first with CN8AW, HP9ABX, K8YWC, UA-1, 3, OX4BK, VE1UW, W-1KVV, 2CPO, 2ZPO, 3HUM, 4WK, 9CJZ, ZB2-AJ, AM, ZL2BE (830 hrs), while on 40 the same set up raised DJ, DL, DM, F, GD, LA, LZ, OZ, T4ZUF (?). From further out—JA2BAY, KG4NV, VK1ATU.

Twenty and Up

More a daytime band but still a hive of activity both phone and c.w. as the following SWL's report. **I. Black** (Gillingham), HE3O, W6BCX Multee antenna, FR7ZC, H11JMF, KZ5LC, MP4BCC, OA4KY, OD5-BV, BZ, EN, PZ1BW, UI8AE, VK-2SG, 3LG, 3VJ, 3ALB, 3ATO, 5RT, 6RU, VP2AA (Antigua), V5FRB, VP6KL, XE3MF, ZB2AJ, ZL2-UW, BG, 4X4AS, 5A1TZ, 7Q7PBD, 7X2MD, 9G1TF, 9J2AB, 9M4TX. **C. Clayton** (Fife), 840c, 60ft. 1-wire, KV4CI, KR6MM (Okinawa), CN8MH, CO2CO, CR6JA, CR7IZ, EA9AQ, LU6FA, OX3LP, UA9KDK, U18ID, UM8KAK, VK3-YS, AXK, VK6WT (Perth), VQ8BJ, W6-EBG, LVF, VS9ARU, ZD7IM, ZD9BE, ZS5UP, ZS6J. From Chepstow one "Colin" complete with 888 and 132ft. 1-wire heard CR7-BV, CZ, KP4NN, KV4CX SV1CX, UA6LP, VK6DR, VS9AWL, ZB4BCA, ZC4LK, ZE6JL, ZS6XP, 4X4IK, 9G1FL, **D. Wraige** (Wirral), AR88-LF and 15-20ft. indoor ant. managed to hear VK3VS, W6AM and W6UED on s.s.b., while **Chris Peel** (Stoke-on-Trent), S750, 90ft. 1-wire around the loft, went fishing for DX. Cream of the catch include CN8FH, CP8AU, CR6AJ, CX4AW, ET3USA, FK8PH, HK7KD, JA2NA, KG6API, KR6CH, KL7BIC, KZ5AW, LU7FAG, OA4KY, PY1FK, PY5AM, TF2WHI, UA9KFS, VE7MS, VE1AED/SU, VK-2ADA, GW, 3ADR, UK, 4SD, 5SM, VP1CY, VP5RB, VU2CK (India), VQ8BFA, W1, 2, 3, 4, 5, 8, 9, Ø, W6-ILT, ABC, AYM, IPA, W7-SEG, QBA, MKI, XW8AX, YA1AW, YA3LCC, YN4CM, YV4IM, ZC4MO, ZL-2UW, 3NO, 4BY, ZS6ALM, 5ASTA, 9Q1FS. While on 15 metres Chris logged CR4BC, CX1WG, EA7GK, ET3USA, HK4XP, JA6QT, H18XM, OA4KY, OX3JV, PY2DXI, VS9AWR, W's, XW8NZ, 4X4UG, 5A4TI, 5A5PJ, 5N2FEL, 5Z4AA, 7Q7BN, 9E1VF, 9G1RM, 9H1R, 9K2AD, 9U51B, 9Q5QR. **F. Simpson** (Hull), RX80, 1-wire, exact length unspecified, reeled in CR7BL, CN8FT, EA8DV, EP3RO, KP4CNC, KW6EJ, KV4CX, MP4TBO, OA8V, VE2AFJ, VK-2NN, 3JA, 4RO, 9PL, VS9AWR, XW8AZ, ZE1AN, ZS1CZ, 4X4SC, 7Q7BN, 9H1S, all on 15 metres. On 20 the same length of wire heard DU1AA, FK8AC, JA1MJ, KA5RC, KR6LU, KW6EJ, KX6DR, PY2BYU, PZ1BW, VK3UQ, VP2AA, VS6AJ, YA1AW, ZL3AB, 9G1LS, 9L1HX. Reports on 28Mc/s are still coming in. **C. Clarke** (Farnham), 12 valve s/het, dipole, G's by the score plus UP2ADZ, ZE2JA, ZS9G, mostly a.m. with a few ssb. **F. Simpson** again heard HB9ED, IICGV, SVIDZ, ZE1AN, ZE2JA, ZE3JU.

Finally a surprise, a genuine log for 144Mc/s, our very first. The honours for the christening go to **David Douglas** (Dundee), who says the following were "got" with an HRO. (With a converter, I hope), G3-BRA 80, HUI 200, MCR 200, HBØLL, 750, SV1AB 1,200, YU1EXY 750. Numbers refer to distance from the transmitting stations in miles.

BOOKS REVIEWED

BASIC ELECTRICITY/ELECTRONICS LABORATORY WORKBOOK. By Training & Retraining Inc. Published by W. Foulsham & Co., Ltd., Slough, Bucks. 224 pages. Size 10½ x 8½in. Price 35s.

THIS is a book whose introduction was printed in England and the balance in the U.S.A. Doubtless in America this is a useful publication but your reviewer doubts its popularity this side of the Atlantic.

It presents a number of projects involving the use of a VTVM and/or oscilloscope. Each chapter is based on a question and answer technique so that the reader is quizzed on knowledge gained as the book proceeds; the answers are given later in the chapter. Projects range from a simple power supply to a superhet receiver but, as one proceeds, various doubts tend to arise.

Constant irritating references to other volumes (by the same publishers perchance) are one source of doubt. Another is the number of odd things which do not seem to tie up. Page 119 assures that we will need a 1,500Ω resistor, yet the relevant circuit on page 122 depicts 1,500kΩ. The availability of a three-gang 365+365+162pF also might prove difficult. Verdict: Go to America to read this one.—*DLG*.

TAPE RECORDING SERVICING MANUAL. By H. W. Hellyer. Published by George Newnes Ltd. 336 pages. Size 9½ x 7½in. Price 33s.

ON first flicking through the pages of this monumental tome the reader is likely to register surprised anticipation; on closer investigation he might well exclaim: "At last!" For while there have been, and still are, books on tape recorders this one reaches a degree of completeness never before achieved. The trouble with most books is that excellent though they may be in themselves, they try to cover too wide a field in the space available. Those interested in the servicing aspects often have to make shift with an odd chapter or two. Here at last is a book devoted exclusively to tape recorder servicing.

The author, H. W. Hellyer, will need no introduction to readers of *PRACTICAL WIRELESS* and *Practical Television*, but it is worth pointing out that in producing his *magnum opus* he was able to draw on his own considerable practical workshop experience of the wayward behaviour of tape equipment.

The introductory chapter covers only 14 of the total 336 pages but condenses in that space a lot of useful information on the basic principles of tape recording. The rest of the book is devoted to an analysis of tape recorders under 62 different trade names! A typical entry details the most important aspects of the machine in question—specification, special features, dismantling, together with notes on special adjustments and servicing with pointers to particular faults peculiar to that model or range of tape recorders. Much of the practical material as mentioned before, is the result of personal experience.

In this way not far short of 300 different tape recorders or decks are dealt with. Circuit diagrams accompany each model or range with mechanical details and component layouts where they serve a useful purpose. In so far as a book of this kind can ever be complete this one is.

Your reviewer has only one criticism and that is the drawings. They are taken from manufacturers' literature and are therefore in almost as many styles as there are circuits, which gives the production a somewhat untidy aspect. However, the meat is there and no doubt the economy involved in not redrawing everything to a standard is reflected in the cost of the book. Still, a pity.

Anyone with any interest at all in tape recording from the servicing angle will miss the buy of the year if he doesn't order this book. For, taking into account the wealth of accumulated data and the potential value of the material, it would be false economy indeed not to add this work to his bookshelf!—*WNS*.

TRANSISTOR RECEIVERS AND AMPLIFIERS. By F. G. Rayer. Published by Focal Press Ltd. 168 pages. Size 8½ x 7½in. Price 30s.

WITH the increasing cheapness and availability of transistors there seems little doubt that the number of enthusiasts using them will also increase. If you are one of these people and you require a not too technical book on the subject of receivers and amplifiers then your problems are solved for 30s. Commencing with an explanation of how these devices work the book continues with chapters on aerials and r.f. amplifiers, mixers, i.f.'s, detectors and a.g.c., Class A, Class B, high-power and hi-fi stages. Bonus chapters include v.h.f. equipment, record players, and test equipment and fault finding. A useful book indeed, steering a practical middle course between high technicalities and boring simplicity.—*DLG*.

MUSICAL INSTRUMENTS AND AUDIO. By G. A. Briggs. Published by Wharfedale Wireless Works Ltd., Idle, Bradford, Yorks. 238 pages. Size 8½ x 5½in. Price 32/6d.

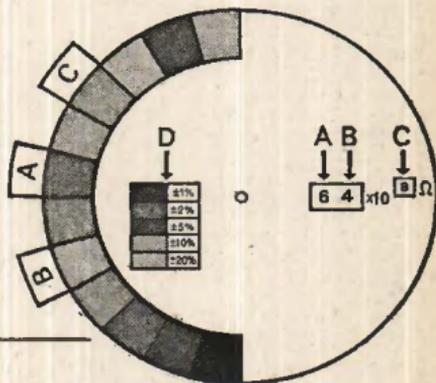
THIS book attempts to bridge the concert-goer and the audiophile both at the same time. The net result is that it falls between two stools, and at 32s. 6d. it is questionable whether it is worth bending down to pick it up. It discusses instruments of all classes, their manufacture, the sounds they emit, wave forms, the formants, harmonics and just about everything. It is degraded by frequent attempts by the author to be funny, and this is further aggravated by the numerous unfunny cartoons and sketches from old issues of *Punch*. There are special chapters written by experts in their field, and had the general style of the book followed the examples set by these writers then the improvement would have been such as to justify a firm recommendation, as it is, however, your reviewer remains unimpressed.—*DLG*.

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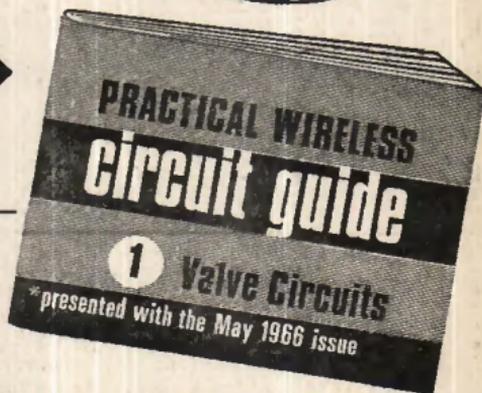


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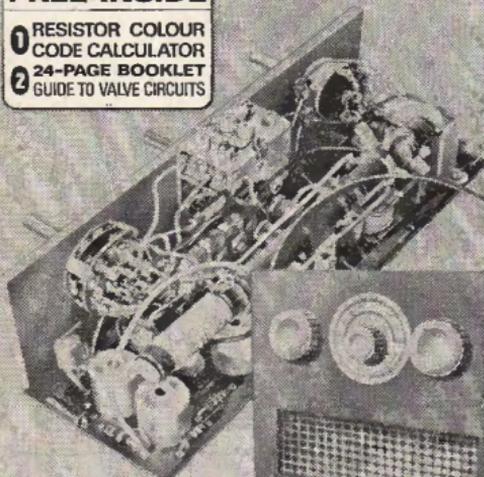
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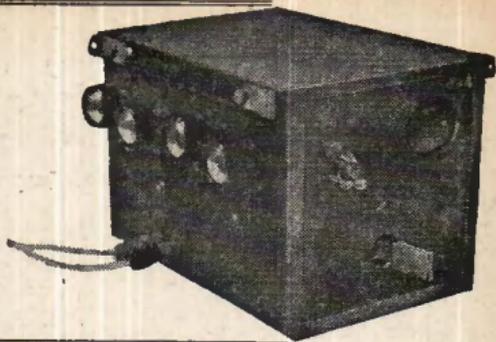
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EL KEY

ELECTRONIC MORSE KEY

P. MURPHY



THIS device was built, taking into account the large variety of types of morse one meets with in the communications world. Whether it be slow or fast characteristic or otherwise, e.g. slow dashes—fast dots, or fast dashes—slow dots, short "space" or long "space", and not forgetting of course perfect well balanced morse. The key was thus designed to be as versatile as possible to take into account the above mentioned types of morse, while retaining the advantage of being an automatic device, and works over a range varying from a very very slow pace up to about 40 w.p.m. The key is therefore suitable for both amateur and professional and would be a very useful asset in training establishments. The key is inexpensive to make, suitable for the home constructor to build and has the advantage of being portable. It is economical to run and uses standard transistor batteries.

The newcomer to the electronic key may at first think it is difficult to use, but like everything else a little practice each day for a week or so will soon convince him that it is not as difficult as it first seemed.

Construction

Handle or Paddle:—The construction of the handle can present a small problem, as it must be capable of moving from side to side and returning to its neutral or "central" position automatically when the finger pressure is released from it, also the gap between its central position and the dot or dash stud into which it comes into contact must be very small indeed, approximately between $\frac{1}{8}$ and $\frac{1}{16}$ in., and a general rule which can apply here is that it is better to have too small than too large a gap.

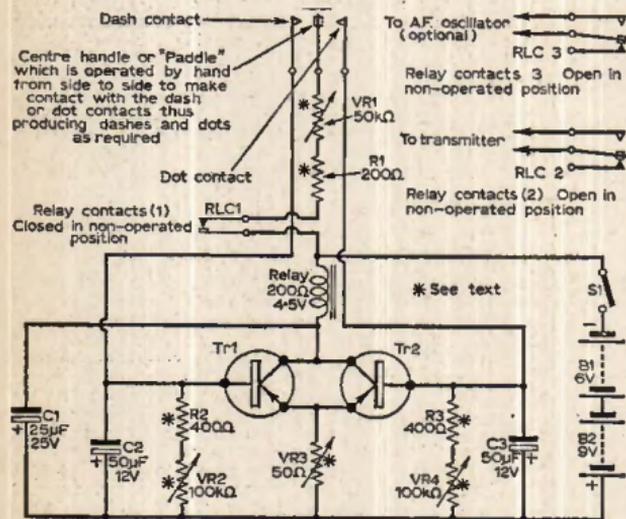


Fig. 1: Circuit diagram of transistorised key.

The paddle described, if constructed properly, can give excellent results and is both simple and easy to construct. However the would-be constructor should pay particular attention to the paddle, a bad handle can spoil the performance considerably, e.g. the letter C is sent thus: — — — a handle which has, for example, a large gap can give an effect which could sound like — — — which produces the letter N twice, an experienced electronic key operator can, of course, overcome things like this, but for the beginner a good paddle is most essential. See Fig. 3.

The paddle consists of a nail file (5in. long), a piece of wood $3\frac{1}{2}$ in. x $1\frac{1}{2}$ in. x $\frac{3}{4}$ in. upon which the nail file is mounted, and an L shaped piece of metal $1\frac{1}{5}$ in. long and $\frac{3}{4}$ in. wide in the horizontal and $1\frac{1}{5}$ in. x $\frac{3}{4}$ in. in the vertical plane (See Fig. 3) which is used to support the nail file on the wooden base. The L shaped piece is secured to the nail file by small nuts and bolts.

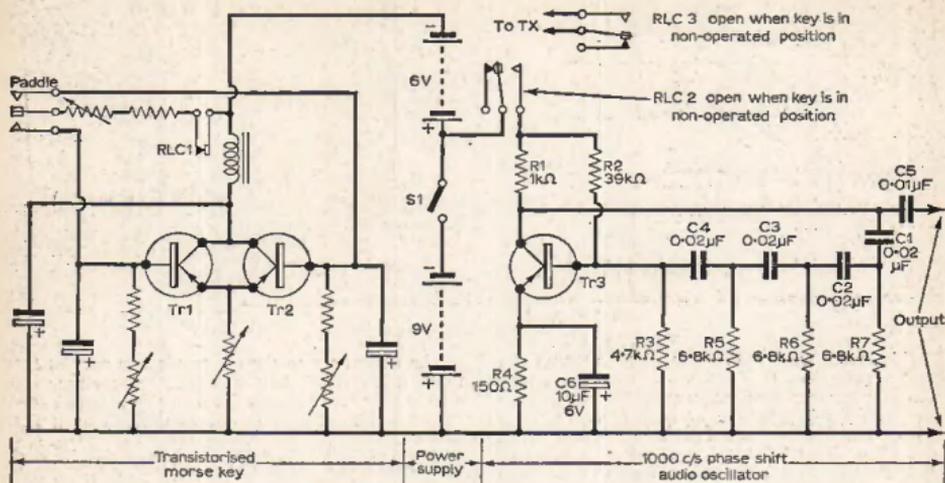


Fig. 2: Diagram of electronic key and phase-shift audio oscillator.

Two flat pieces of formica can be glued to each side of the nail file near its end, which tends to make the paddle easier to operate.

The wooden base can be either glued or screwed to the chassis.

Function of Components in Fig. 1

The components, VR1, R1, C1, make up the time constant (CR) for the space (time interval) between consecutive dots or dashes. Increasing VR1 will therefore increase the space, or vice-versa, the max. space limit chosen was the max. practical limit that would be required although spaces up to unity can be obtained with this type of transistor. They are not required in this case. The min. space depends approximately on the value of C1. In the circuit (Fig. 1) extreme max. and min. values should be obtained, and here as well as the other

variables in the circuit, the constructor may, if he so desires, choose closer limits by using the appropriate value of VR1, which he can quickly determine by experiment. The mark length is slightly affected when varying the space control VR1.

RI Under no circumstances must this 200Ω resistor be omitted or reduced in value as it is a limiting resistor and 200Ω was found to be the min. critical value.

The components, VR2, R2, C2, make up the time constant for the dash side of the circuit. C2 and R2 by themselves provide for the shortest length of dash and by increasing VR2 the length of dash can be increased, here again max. and min. values have been chosen, and it is up to the constructor if he wishes to modify these.

The components, VR4, R3, C3, make up the time constant (CR) for determining the length of dot and as in the case of the dashes extreme max. and min. values were again chosen, R3 and C3 together decide the shortest length of dot and VR4 can be increased to lengthen it.

VR3 varies the overall speed, the ratio of dot to dash and space etc., are fixed by VR1, VR2 and VR4, and VR3 varies the speed of the cycle as a whole, though slight variations do occur in the ratios especially at higher and lower speeds. However this is easily overcome by setting the ratios at approximately the speed one is going to work at and adjusting as necessary. In any case only one control will need adjusting. Increasing VR3 decreases the speed and vice versa.

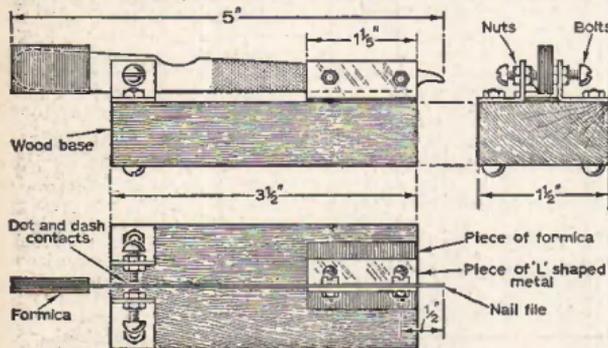


Fig. 3: Complete "paddle" for the key.

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OC2	5/8	9B8T	10/6	6L1	8/6	10L11	10/6	50C15	9/6	ATP4	1/8	RB91	3/6	EP08	10/6	N78	15/6	FY8	8/6	UCL88	8/6
OC3	5/8	9B8S	7/6	8L5Q	7/6	10P18	12/6	30C17	12/6	ATP5	7/6	RB03	6/6	EP158	6/6	N106	15/6	PV800	5/8	UF41	7/6
LA7	7/6	9B8T	16/6	8L1E	7/6	12A8H	9/6	30P3	8/6	ATP7	4/6	EB241	6/6	EP184	6/6	NG21	8/6	FY801	6/6	UF68	8/6
LD5	6/6	9BWS	8/6	8Q7G	8/6	12A7T	8/6	30L15	18/6	AU3	8/6	EB030	3/6	EL32	5/6	NG27	8/6	RL	4/6	UL41	6/6
1R6	7/6	9BW7	8/6	9Q7GT	9/6	12A7T	8/6	30L15	18/6	AU5	6/6	EBP80	5/6	EL38	17/6	OZ4	4/6	R19	7/6	UL84	6/6
LLD8	5/6	8C4	1/6	98A7	4/6	12A1V	4/6	30L17	12/6	AZ1	8/6	EBP88	3/6	EL34	9/6	PC88	10/6	RG5/80	8/6	UM80	7/6
1N50T	8/6	8C0G	4/6	98C7	6/6	12A17	5/6	30P18	10/6	AZ31	7/6	EBP89	6/6	EL41	7/6	PC88	9/6	R41	9/6	UC15	13/6
1R5	4/6	8C8	8/6	98C7	4/6	12A1X	4/6	30P19	14/6	CB181	8/6	EBL1	17/6	EL42	7/6	PC87	7/6	8180	10/6	UU7	10/6
1R4	5/6	8C80	8/6	98H7	2/6	12B4E	6/6	30P11	11/6	CK303	5/6	EBL21	10/6	EL54	4/6	PC84	5/6	8P4	9/6	UU9	8/6
1B5	3/6	8C80G	22/6	98J7	3/6	12B8	4/6	30P18	12/6	CL31	18/6	EBL31	27/6	EL80	6/6	PC85	8/6	8P4	1/6	UY21	7/6
1Y4	2/6	8C84	3/6	98K7GT	4/6	12B8T	2/6	30P114	12/6	CM3	18/6	ECU81	3/6	EL85	8/6	PC169	10/6	8P61	1/6	UY41	4/6
5A4	3/6	8CWA	14/6	98L7GT	4/6	12C8GT	7/6	38A5	17/6	DAC32	7/6	ECU82	3/6	EM34	9/6	PCF80	8/6	8U22	19/6	UY85	4/6
3Q4	6/6	8D6	8/6	98N7GT	4/6	12E1	19/6	38L5	5/6	DAF91	3/6	ECU88	4/6	EM60	8/6	PCF82	4/6	8U2180	12/6	VCR8	4/6
8Q5	6/6	8E5	5/6	98Q7	6/6	12J8GT	3/6	38W4	4/6	DAF96	6/6	ECU94	6/6	EM71	7/6	PCP84	8/6	R41	9/6	VCR8175	8/6
8V4	4/6	8F1	9/6	9I44GT	10/6	12J7GT	7/6	38Z3	10/6	DDC80	8/6	ECU88	5/6	EM81	8/6	PCP85	8/6	TD4	7/6	80/	80/
8T4	8/6	8F6G	5/6	9I59G	7/6	12K7GT	5/6	38Z4GT	3/6	DF82	8/6	ECU88	8/6	EM15082/6	PCF81	9/6	TH41	8/6	7/6	VCR817C	80/
8E40	3/6	8F6G	4/6	8V3M	8/6	12K8GT	8/6	38Z5	5/6	DF70	5/6	ECU88	8/6	EY31	8/6	PCF82	9/6	U10	7/6	VMP4G	17/6
8V40	4/6	8F6G	4/6	8V6G	8/6	12Q7GT	3/6	37	5/6	DF91	2/6	ECU82	6/6	EY66	6/6	PCF80/10/6	U14	7/6	80/	VMP4G	17/6
8V40	3/6	8F11	12/6	8V6GT	7/6	128A7	8/6	48	4/6	DF92	8/6	ECU81	10/6	EZ35	4/6	PCF808/12/6	U19	80/	80/	VMP4G	12/6
8Y30T	4/6	8F18	5/6	8X4	3/6	12807	8/6	50B5	8/6	DF95	22/6	ECU83	10/6	EZ40	5/6	PCF805/12/6	U25	18/6	80/	VMP4G	12/6
8Z4GT	8/6	8F14	12/6	8X5Q	4/6	12817	2/6	50C23	6/6	DI77	3/6	ECU82	8/6	EZ80	8/6	PCF808/12/6	U36	10/6	80/	VMP4G	12/6
8Y80L2	10/6	8F88	8/6	8X5GT	7/6	12847	2/6	50CDB98/4/6	DK92	7/6	ECU81	5/6	EZ80	5/6	PC183	8/6	U78	8/6	VT35	20/6	
8A7	15/6	9G6	2/6	7B6	11/6	128K7	2/6	80L6GT	8/6	DK91	5/6	ECU88	6/6	EZ81	8/6	PC184	7/6	U91	11/6	VT91	50/6
8A6G	12/6	8H8	1/6	7B7	7/6	128K7	5/6	75	8/6	DK92	7/6	ECU88	6/6	EZ82	8/6	PC185	8/6	U92	10/6	VT91	50/6
8A6T	3/6	8J8M	8/6	7C8	10/6	128K7	20/6	78	4/6	DK93	6/6	ECU88	6/6	EZ82	8/6	PC186	8/6	U91	12/6	VT91	50/6
8AK5	4/6	8J8G	2/6	7C8	6/6	19AQ6	7/6	80	5/6	DL70	7/6	EAL83	9/6	EZ84	8/6	PEN44	80/	U92	6/6	VY88	25/6
8AL5	3/6	8J8GT	4/6	7D5	8/6	30D1	10/6	86A3	8/6	DL82	4/6	ECU86	8/6	EZ80	22/6	PEN44	20/6	U91	18/6	VY88	25/6
8AMS	2/6	8J8	3/6	7D7	8/6	20P2	11/6	150B8	14/6	DL83	3/6	EP3	17/6	EZ80	8/6	LAC83	8/6	X78	20/6	VY88	25/6
8A8S	2/6	8J8M	8/6	7D7	12/6	20L1	15/6	130C4	12/6	DL84	6/6	EP8	8/6	ET06	20/6	PEN46	2/6	CAF24	7/6	X78	41/6
8AMS	0/6	8J7G	4/6	787	18/6	20P4	14/6	801	5/6	DL85	6/6	EP7A	2/6	ET11	10/6	PL36	9/6	UB41	6/6	XH1-5	41/6
8A8T	22/6	8J7GT	8/6	7Y4	3/6	30P5	12/6	807	7/6	DL86	6/6	EP8	2/6	ET85	25/6	PL33	3/6	UR91	7/6	XPI-5	5/6
8AT5	3/6	8J8GT	5/6	8W8	8/6	28A5	6/6	818	20/6	DM70	5/6	EP41	6/6	ETW61	4/6	PL82	5/6	URF80	6/6	X81-10	5/6
8AT8	0/6	8K7M	5/6	10C1	12/6	32L6GT	8/6	868	10/6	DY88	7/6	EP40	2/6	ET241	8/6	PL83	8/6	URP89	6/6	Y88	7/6
8B6G	2/6	8K70	1/6	10C2	12/6	28Y8	8/6	954	4/6	DY97	7/6	EP80	4/6	ML4	17/6	PL84	9/6	UC88	8/6	ZB4	40/6
8B4S	4/6	8K7GT	4/6	10T1	12/6	32Z4	8/6	1888	5/6	ES8C0	14/6	EP86	4/6	MX40	12/6	PY31	5/6	UC88	8/6	ZB7	12/6
8B8H	7/6	8K8G	8/6	10F3	12/6	28Z5	7/6	402AR	15/6	EA00	8/6	EP86	4/6	MX44	4/6	PY33	8/6	UC82	7/6	ZB7	12/6
8B8H	7/6	8K8GT	3/6	10F18	9/6	33D7	5/6	7183	1/6	EAF43	7/6	EP81	2/6	MX40	12/6	PY31	5/6	UC81	6/6	ZB7	12/6
8B70	7/6	8K8GT	3/6	10F18	9/6	33D7	5/6	7183	1/6	EAF43	7/6	EP82	2/6	MX44	4/6	PY32	8/6	UC82	7/6	ZB7	12/6
8BQ7A	7/6	8K25	20/6	10L4	10/6	30C1	9/6	7472	2/6	EB41	6/6	EP82	2/6	MX44	4/6	PY32	8/6	UC82	7/6	ZB7	12/6

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Transistors

These are GEC S3 type transistors. Type GEC S1 will perform equally well and although others have not been tried in Fig. 1, it is presumed that any p-n-p type transistor which is suitable for a class B output stage e.g. Mullard OC72 would be suitable. GEC types were chosen because they were less expensive at the time. It may be necessary to adjust some of the component values, if other types of transistors are used, because differences in performance was noticed, even with individual transistors of the same type, but these variations were slight.

Layout and General Information

For the right-handed operator it is necessary that all the adjustable controls can be easily turned with the left hand, so that adjustments can be made whilst actually operating, and saves the necessity of having to stop to alter the ratios or speed etc. A left hand operator should have access to the variable controls with his right hand.

The two pairs of wires from the relay contacts RLC2, RLC3 should each preferably be terminated in a plug and socket at the back of the box or chassis to prevent wires going to the tx and osc. (only for separate osc. not built in the key as in Fig. 1) from getting in the way near the paddle. The on/off switch would naturally be convenient in the front, (when Fig. 2 is built only one pair of contacts need be terminated in plug and socket for tx). The max. height of the handle above the bottom of the box or chassis should not exceed 2in., this height also includes the width of the nail file.

When operating the key the correct position of the arm and hand is absolutely essential and it is as follows:— The arm from the elbow down should lie relaxed on the table, with the thumb side of the hand upwards, the handle of the key is then operated with thumb and first finger in a gentle side to side movement. The whole of the hand up to the wrist should follow the movement of the fingers and in this way, an operator can send for hour after hour without getting tired. The writer knows of no other method which can equal that just described for operating the key and getting

the best results from it with the minimum of effort.

When assembling the device; especially when experimenting with it, it was found on several occasions that minute pieces of solder, dust, etc., got between the relay armature and pole face preventing the key from functioning, and it was only after testing everything else that the relay was examined and the fault discovered; however, a thorough clean-up after the unit is placed in the box would avoid this fault altogether, together with careful soldering.

When Fig. 2 is built (i.e. key and oscillator together) a loudspeaker may be built-in or a "split phones" arrangement with a 3-way switch, i.e., e.g. Switch Position

1. One earpiece to Radio Receiver, other earpiece to audio oscillator.
2. Both earpieces to Radio Receiver.
3. Both earpieces to oscillator.

While on the subject of Fig. 2, the current consumption with oscillator and key drawing from the batteries, varies between some 20 to 30mA, negligible current is drawn when not keying; however the on/off switch from the supply should be switched off when the key is not in use.

The photograph shows the key when built; it can however be enclosed in a much smaller space, and must not be too light in weight although the batteries do help to make up weight, also the width should not be less than say 3 to 4in. This helps to stabilise the unit and prevent it from moving when keying. Component tag boards were screwed to the side of the box and the opposite side was used for the variable pots. A suitable lid fits on top and the workings are then completely enclosed.

Although there are many different combinations of capacity and resistance which would give the same results as the key in Fig. 1 the circuit design is a standard, and providing the constructor can comply with the transistor limitations, and provide enough current to operate the relay, a wide range of combinations are open to him. Dots are normally to the right operated by the thumb, dashes are normally to the left operated by the first finger. In Fig. 1 both dots and dashes can be obtained from either side, should an operator require the opposite.

ON THE SHORT WAVES

—continued from page 1043

There are VK's on topband and at least one of our reporters has heard a JA. Anyone else hear these remote parts at this frequency? For those wanting to know when to listen on 20 and 15 the answer is that they are open from 0600-1800 G.M.T. Someone will doubtless hear a piece of fabulous DX at 1081 now that I've said that! Listen on phone only for VQ8AZ and VQ8BZ in Mauritius. ZD8's are planning some DX-peditions in the Caribbean area soon, ZD5M is loose on 21 c.w. (Swaziland), while on 20 c.w. TL8SW is quite a rare bird. ZS2MI reported active from Marion Island, and spies report that an expedition is promised for Rio de Oro soon. Contests for March include 12-13th, ARRL DX Contest (phone); 19-20th, BERU; 26-27th, ARRL DX Contest (c.w.); April 3rd, Low Power (QRP) Contest. All logs welcome, deadline for May issue is March 27th.

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PRACTICALLY WIRELESS

THE perennial subject of breakdown has haunted the electronics industry from the "shed-in-the-back-garden" days to this era of chrome and plastic fabricatories, despite the analysis and study of a whole new group of pseudo-scientists, whose main function seems to be to evaluate and correlate and argue at length with each other in the learned journals.

It does not haunt those of us who were raised on the bread-board and makeshift component. Breakdowns were the order of the day. We righteously asserted that one could only learn by one's mistakes. But to a later generation of experimenters and constructors, brought up on ready-made receivers, and elegantly annotated kits, failure is taboo.

It may seem strange that a manufacturer should even admit the possibility of failure when launching a product. Let alone publish his estimate of Mean Time Between Failures, which is what the electronics industry understands by those initials.

The aforesaid pseudo-scientists have great fun working out the MTBF and juggling the figures as dextrously as a radio retailer trying to explain battery working hours to the rather dim purchaser of a portable tape recorder.

It is becoming a modern trend



Trying to explain to a rather dim purchaser.

to admit not only that the equipment can go wrong, but that it will predictably continue in its back-sliding ways. Our postbag shows that many readers are not yet educated into this way of thinking. Any practising service engineer could soon offer a host of demonstrations.

We do not have to look much farther than the cheaper imported transistor radios. Except that a request for repair will show that the MTBF is either zero or infinity depending on whether the harassed salesman is sucker enough to accept the job, or shuns any responsibility.

In the words of the poet: "Radios out of Old Hong Kong, All too frequently go wrong".

And the optimists who imagine that things will improve are due to be disillusioned. Already, rising labour rates, are worrying the Japanese, and have brought the electronics section of Hong Kong's labour force into the "elite" class. With Taiwan emerging as a healthy radio manufacturing centre and the great labour markets of Korea and India as yet untouched it seems likely we shall get more and more of the cheap and nasty cut-price models in the near future.

To digress: a Japanese industrialist told us the interesting story of an argument with a taxi-driver whose radio blared annoyingly in the cab. He took a lot of convincing that Hong Kong was part of the wrong Empire. The sun will never set on shoddy goods—that's for sure! Of course, it is unfair to lay the whole blame abroad. There are too many examples nearer home.

Henry has been taken to task, and is unrepentant, for dwelling on the maker's obsession with profit, to the apparent exclusion of reliability. Apart from the top-quality goods, where competition is for specifications and facilities and where high price seems an added attraction, most development seems to consist of finding



An argument with a taxi driver.

ways of cutting costs.

A nodding acquaintance with the term "serviceability" comes only from the copy-writing lads, who see a hinged chassis or plug in parts and go delirious in print. Original reason for the hinge was probably that it fitted a factory jig or facilitated swifter inspection. The plug and socket connectors are now mass-produced by astute experts who have a formidable range of such products, and can thus supply more cheaply than the set-maker can get his joints laid in and soldered.

There is a notorious television receiver with both these "advantages". Only trouble is that to take full advantage of the hinge it is necessary to partially dismantle the tuner unit and 405/625 system switch, and the first time the serviceman tries it he generally manages to break the edge of the printed board. On later marks, the plug sections have wiring added to the back pins, compounding the felony. The l.o.t. is almost impossible to remove unless the printed board is flexed and the system switch rod bent.

MTBF, in this case, is in inverse proportion to the amount of extra time taken by the service engineer in clearing what may have been originally a simple fault.

Probably one of those plug connectors not making properly. Hong Kong fashion.

FOR
THE BEGINNER

A VERSATILE

PREAMPLIFIER

and

TONE CONTROL

by A. S. Ellis



IN the field of audio, need is often felt for an unspecialised preamplifier capable of being used with a wide variety of equipment for various purposes not necessarily forming a permanent part of any of these pieces of equipment. The particular preamplifier to be described was originally built to give greater sensitivity to a somewhat insensitive "gram" amplifier incorporated in a radio receiver. The preamplifier unit has, however, since then been used with enormous success in a number of other applications.

In the particular "gram" amplifier modified no variable tone controls were present and reproduction of gramophone records tended to be somewhat disappointing at times. Most annoying of all was an almost complete absence of bass. Accordingly the tone control box was constructed

for insertion between the pick-up and the preamplifier. The results, when all modifications had been completed, were astounding. Since then the tone control box, too, has been put to numerous uses, not the least of which has been that of a tone correction device whilst making tape recordings.

No originality whatsoever is claimed for the circuits used; the preamplifier is of fully conven-

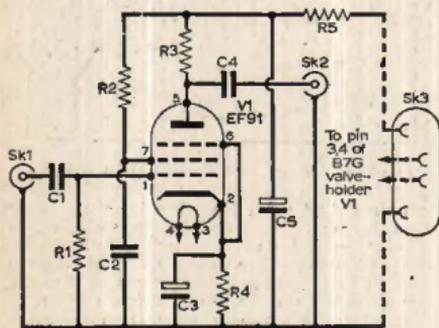


Fig. 1: The circuit diagram of the preamplifier.

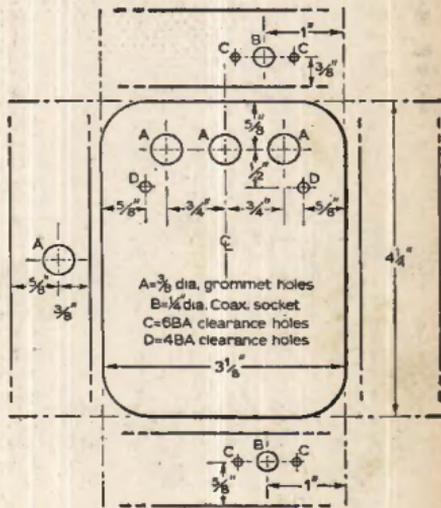


Fig. 2: Preamplifier chassis drilling details (top view).

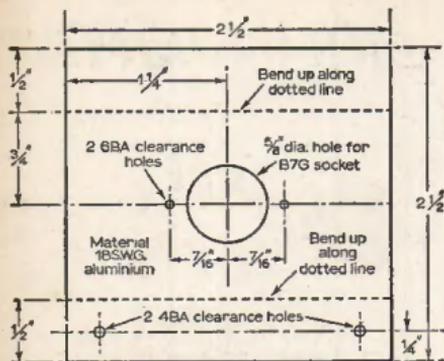


Fig. 2a: Drilling and bending details of valveholder bracket.

tional design, the control circuit is that used in the very popular Mullard two and three valve preamplifiers. However, the form of unit construction is somewhat novel and adds considerably to the versatility of both units.

Externally Powered Preamplifier

The preamplifier circuit is built around the well-known pentode type EF91, which is easily obtainable. In this circuit (Fig. 1) the valve consumes no more than about 1mA h.t. It also requires a heater supply of 6.3V at 0.3A, but both these requirements should easily be met without any overloading of the power-pack of the equipment to which it is attached.

Construction of this unit was carried out in and around a 2oz. tobacco tin of the variety measuring

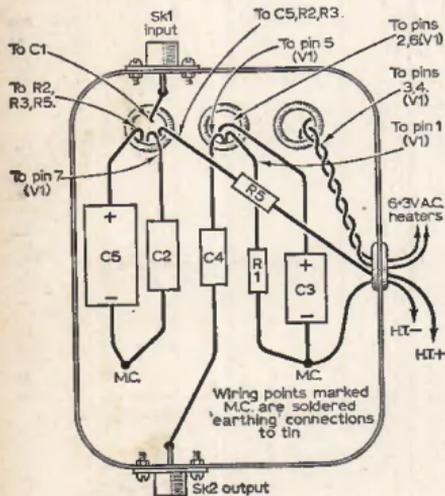


Fig. 3: Preamplifier component layout and wiring.

PRE-AMPLIFIER COMPONENTS LIST (Fig. 1)

Resistors:			
R1	1M Ω	R4	2.7k Ω
R2	1M Ω	R5	15k Ω
R3	220k Ω		
All 10% $\frac{1}{2}$ W carbon.			
Capacitors:			
C1	0.1 μ F	paper	150V.
C2	0.1 μ F	paper	350V.
C3	50 μ F	electrolytic	12V.
C4	0.1 μ F	paper	350V.
C5	16 μ F	electrolytic	350V.

Miscellaneous:

V1, EF91, SK1, SK2 coaxial sockets. PL1 power supply plug to fit socket mounted on receiver or other equipment, type optional. 2oz. tobacco-tin, size approx. 4 x 3 x 1 in.

Small piece aluminium $2\frac{1}{2}$ x $2\frac{1}{2}$ in. One B7G valveholder. Nuts and bolts.

TONE NETWORK COMPONENTS LIST (Fig. 4)

Resistors:			
R1	47k Ω	R3	68k Ω
R2	39k Ω	R4	6.8k Ω
All 10%, $\frac{1}{2}$ w.			
Capacitors:			
C1	560pF	silvered mica	
C2	8200pF	silvered mica	
C3	2200pF	silvered mica	
C4	0.02 μ F	paper	150V.

Potentiometers:

VR1	250k Ω	log.	treble
VR2	250k Ω	log.	bass
VR3	250k Ω	log.	volume

Miscellaneous:

2oz. tobacco tin, approx. 4 x 3 x 1 in., 3 miniature instrument knobs, SK1, SK2 coaxial sockets.

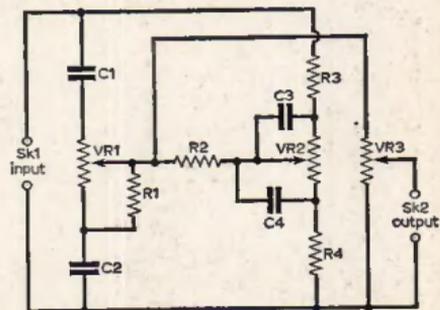


Fig. 4: Circuit diagram of tone control network.

approximately 4 x 3 x 1 in. The lid of the tin was removed, later to become the baseplate of the preamplifier. The tin itself was then turned upside down and used as a chassis. Drilling details are shown in Fig. 2. It was decided not to mount the valve vertically on the tin as it was thought that wiring beneath would become unnecessarily crowded and inconveniently placed. Accordingly a piece of aluminium $2\frac{1}{2}$ x $2\frac{1}{2}$ in. x 18 s.w.g. was

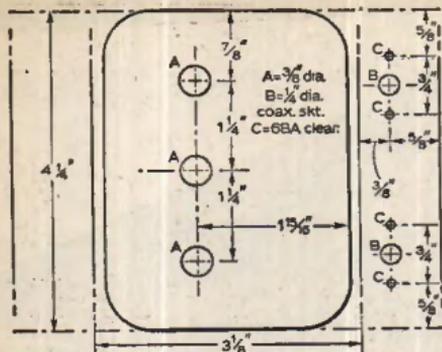


Fig. 5: Tone control chassis drilling details.

cut, drilled and bent as shown in Fig. 2a so that the valve could be mounted horizontally. This support was secured to the main chassis by means of 4BA nuts, bolts and shakeproof washers. Wires to the valveholder were taken from inside the tin via holes A, B and C (see Fig. 2). (Heater wiring was kept well away from the rest of the circuit.) Small components such as R1, R2 and R3 were wired directly to the valveholder but the remainder of the components were mounted inside the tin out of sight (see Fig. 3 underchassis point-to-point wiring diagram). Earthing connections were easily made by soldering directly to the tin chassis. Power supply leads were taken through hole D1, which was fitted with a rubber grommet. The replaced lid forming a baseplate, completed the preamplifier. *Note:* If desired a volume control may be incorporated in the unit by substituting a miniature 1M Ω log. potentiometer for R1. This can be mounted vertically in the tin after drilling a suitable hole and can be fitted with a miniature instrument knob.

Preamplifier Power Supplies

The unit appears to be fairly tolerant of power supplies applied to it, though an h.t. supply voltage of between 200 and 300V d.c. is preferable. In the prototype the power supply leads were terminated in an old Mazda octal valve base which took on the role of a supply plug. A similarly wired

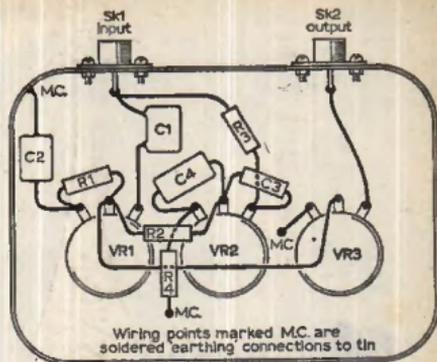


Fig. 6: Tone control wiring and component layout.

matching valveholder was mounted on the main equipment.

Tone Control Box

A very wide range of tone control is available with this circuit (Fig. 4): VR1 provides treble cut and boost; VR2 provides bass cut and boost; VR3 is a volume control. It must, of course, be borne in mind that as with all passive tone control networks there is a considerable insertion loss. However, if the unit is used with the preamplifier already described a fair excess of signal is still available.

As before, a 2oz. tobacco tin was used as a chassis, the lid being temporarily removed to be used as a baseplate later. The drilling details of the tin are given in Fig. 5; holes A, B and C are for the bushes of the vertically mounted potentiometers. (It is recommended that miniature potentiometers be used, otherwise some difficulty may be encountered in fitting all three components into the tin.) There is ample room in the tin for all wiring and components may be supported on the potentiometer and socket connections (see Fig. 6). When complete and with lid replaced the circuit is completely screened; there will therefore be no hum problems.

No difficulty should be experienced in building either of these units even by the newcomer to audio circuitry. ■

PRACTICAL TELEVISION—APRIL

L-C PICOMETER

used with a G.D.O. this instrument enables you to measure small capacitance and inductance values accurately and speedily.

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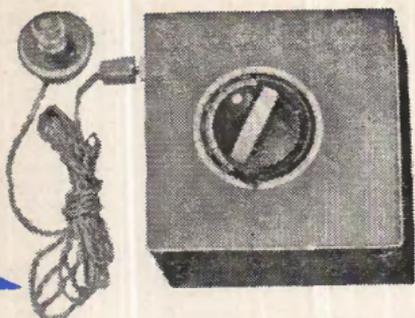
FAULTY CHOKES AND TRANSFORMERS — A look at the situations where these are to blame for the trouble.

TV TERMS AND DEFINITIONS

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On sale 24th March 2/-

X
2



by the reflexed action is well worth while in small receivers.

Now, the audio signal at the junction of R2 L2 is applied to the base of Tr2 through the electrolytic coupling capacitor C4. Tr2 is arranged as a low-level second audio amplifier for working the crystal earpiece. The audio signal is developed across the collector load R5 and is directly coupled to the earpiece via the jack socket, at terminals 1 and 2. Terminal 3 on the jack socket serves as a switch, so that a switch contact in the jack makes when the jack plug is inserted, thereby connecting the battery supply circuit from battery negative to the negative line of the circuit, via jack socket terminals 3 and 1.

Switched Jack

Miniature "switched" jacks are available, but those investigated by the author are arranged so that the switch opens when the earpiece jack is inserted. This action can be reversed, so that the switch closes when the jack plug is inserted, by easing the spring switch contact so that it is above the jack contact (contact 3) which is activated by the plug. Fig. 2 shows at (a) the jack socket in its ordinary state and at (b) the socket with the contact rearranged in the manner described above.

Both transistors run with very small emitter current, and this current is determined by the values of R1 (for Tr1) and R4 (for Tr2). R3 and C3 serve to decouple the two stages and improve the stability margin.

Tr1 is a Mullard AF127, which is a smaller ver-

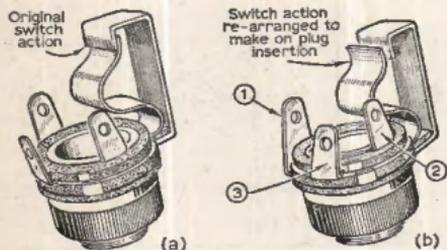


Fig. 2: Modifications for the phone socket.

sion of the AF117. The latter can be used but its larger size takes up most of the room between the tuning gang and the ferrite rod aerial. Tr2 is a medium gain OC71. The diodes D1 and D2 are miniature Mullard OA91's or equivalents. It is important to make sure that they are connected round the right way in the circuit otherwise the sensitivity will be impaired.

The receiver is built on a laminate or bakelite panel of approximately 2½ x 2½ in. Most of the components are secured to the panel and connected in circuit by the use of small eyelets fitted in holes drilled in the panel. In addition to the eyelet holes, the panel carries a ½ in. diameter hole for mounting the tuning gang, two holes for L2 coil former, two holes for the rubber band that fixes around the ferrite rod aerial mounting grommets, two holes for the battery securing rubber band.

Drilling details for the panel are given in Fig. 3, and the holes marked with "Y" are those to accommodate the eyelets, of which there are twenty-four. These holes should be of a size that provides the eyelets with a fairly tight fit, the actual size, however, being determined by the type of eyelet employed. A ready drilled chassis board is available plus twenty-four push-fit eyelets (see components list). A No. 48 drill is used for the former fixing holes of L2, while a ½ in. drill is used for the rubber band holes that are used for securing the ferrite rod grommets and battery.

Fig. 4 shows how the holes in the panel are employed, and this also incorporates a point-to-point wiring diagram. The eyelets, it will be seen, carry the capacitors and resistors on the left-hand side of the panel, also the lead-out wires of Tr1 and Tr2 (the latter in the top left-hand corner) in addition to the two diodes and capacitor C1 (near the ferrite rod aerial). The broken lines on this diagram correspond to point-to-point connections made beneath the panel, while the full-line connections are those made on the top of the panel.

When the holes are drilled in the panel and the eyelets fitted, the next move should be to solder the components to the eyelets, after which the larger components can be more easily fitted. As each component is soldered to the appropriate eyelet the above- or below-panel interconnecting wire or wires should also be soldered. These wire interconnections can either consist of thin, flexible p.v.c. covered stranded wire or about 26 s.w.g. tinned copper wire covered with insulated sleeving. Stranded miniature p.v.c. wire was found to be the best for the job by the author.

The eyelet-connected components on the left-hand side of the panel are mounted vertically, and it is this kind of mounting that enables all the components (of "standard" size) to be accommodated on the board. The idea is to dress one of the lead-out wires from resistor or capacitor back along its length, so that the two ends can then be pushed into the adjacent eyelets. It is a good plan to put a short length of insulated sleeving over the wire running by the side of the component to prevent any possibility of short-circuiting.

A miniature soldering iron will greatly facilitate the various joins and prevent burning the insulated sleeving and components. It is important, though,

continued overleaf

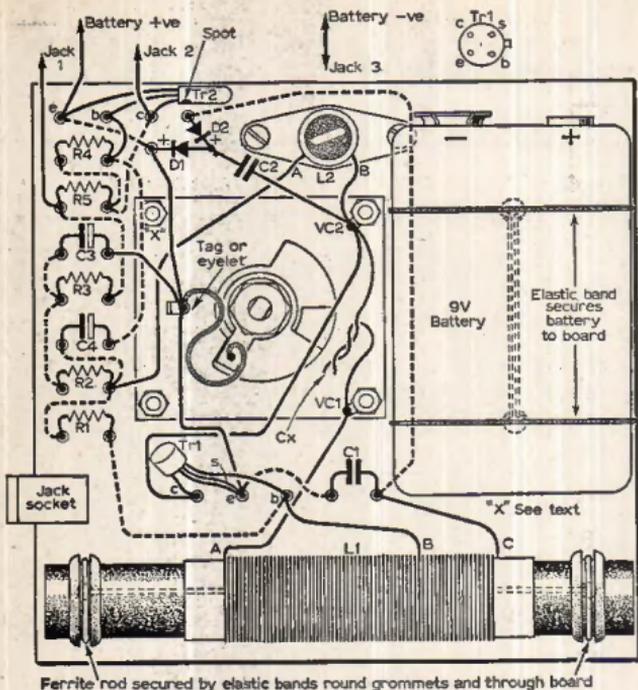


Fig. 4: Component layout and wiring details.

It is best to produce this winding either on a separate former or on the former to be used before it is finally fitted to the panel. The card former can then easily be slid over the former once the latter is in position and screwed down.

The ferrite rod is held clear of the panel by two ¼ in. rubber grommets, one at each end. These are also used for securing the rod a thin rubber band passing round them and through the holes at each side of the panel. A similar rubber band is used to hold the PP3 battery to the panel (see Fig. 3).

Finally, a thin flexible two-conductor lead should be made up for the battery connections, the ends of the conductors terminated in suitable connectors for the PP3 battery and a three-conductor flexible lead should be processed for connecting from the panel at the various

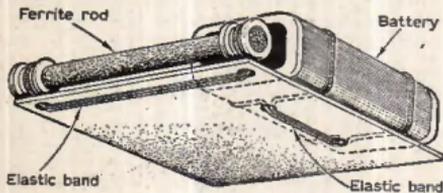


Fig. 5: Method of connecting battery and ferrite rod.

points indicated in Figs. 3 and 4 to the jack socket. Extreme caution should be taken over this latter exercise to ensure that the conductors are terminated to the correct tags or terminals on the jack socket (see Fig. 2). It is also very important, of course, to avoid reversing the battery polarity, for while this may not completely ruin the transistors it could reduce their efficiency and alter their characteristics.

There is sufficient room to accommodate the jack socket between the ferrite rod aerial and R1 (see Fig. 4, for instance), but since the jack socket terminals carry the full negative voltage of the battery, inadvertent contact between one of these terminals and the base end of R1 could immediately destroy Tr1. For this reason, an *insulating sleeve* should be dressed over the terminal end of the jack socket.

Tuning Up

L2 former should be fitted with a dust-iron core and initially this should be adjusted so that it embraces the whole of the winding on the ferrite rod, there should be no difficulty in receiving the local m.w. station. To peak reception on this programme, L2 core should be re-adjusted while slowly turning the tuning gang a little either side of the station for optimum gain.

While it is impossible for a receiver of this kind to track accurately over the whole dial, reasonable tracking is achieved owing mainly to the flat tuning of L2. However, if the set is required to be peaked to a more distant station the station should be tuned as near as possible on the gang and then L2 should be re-adjusted at this new setting for optimum gain.

A degree of feedback occurs in the r.f. stage, especially towards the higher frequency end of the band automatically due to stray capacitance. However increased feedback can be applied simply by flexing a pair of thin, insulated conductors between the two "live" terminals of the tuning gang. This is shown as the "regen coupling" on the circuit (Fig. 1) and the coupling in physical form is clearly shown in Fig. 9. On no account should a d.c. connection exist between the two flexed conductors, for their purpose is simply to provide a small variable capacitance. In the prototype, the five

—continued on page 1081

Modulated Light Telephone

We have experimented with the Modulated Light Telephone Link (February issue) at school and have found results reasonably good although there was a fair amount of background noise and distortion when used outside in daylight.

We discovered that a large lens was better for focusing the light beam onto the photocell than a parabolic reflector. The frequency response at 20ft. went up to nearly 4 kc/s.

I would be interested to know if anybody else tried this and what their results were like.

R. H. Boston.

The Chalet,
Pett Road, Pett,
Hastings,
Sussex.

Copies for Sale

I HAVE the complete copies of PRACTICAL WIRELESS for the years 1959-60-61-62-63-64-65, and if any of your readers are interested I would be willing to dispose of these for a reasonable sum.

All enquiries should be accompanied by a stamped addressed envelope.

W. R. Hebditch.

90 Clarendon Road,
Hove 3,
Sussex.

Who Wants?

OWING to lack of space, I am forced to dispose of an amount of old and new radio chassis and components. Also I have available P.W.'s from January 1961 to March 1965, and a number of various other magazines.

I am willing to give to anyone who cares to call, any of the above items. Sorry, but no letters will be answered.

Nicholas Hansen.

413 Church Road,
Upper Norwood,
London,
S.E.19.

Tapespondent Wanted

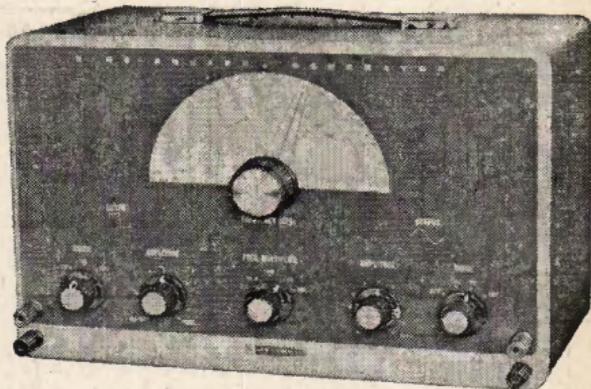
I WOULD like to tapespond with anyone of my own age (16) interested in Hi-Fi radio and popular music.

M. K. Beeny.

28 Wheatfields,
St. Ives,
Huntingdon.

NEWS AND.

SINE-SQUARE WAVE GENERATOR



From Heathkit comes the IG-82U sine-square wave generator which has a frequency range of 20c/s to 1Mc/s in 5 bands. Sine and square wave outputs are available separately or simultaneously and there is less than 0.15 μ Sec. rise time on square wave and less than 0.5% distortion on sine wave. Separate output terminals and attenuators permit individual control. In kit form, the price is £24 10s. and assembled and tested £36 10s.

MULLARD MINIBOOKS

"Principles of Electrostatics" is the title of the first in a new series of "minibooks" announced by the Mullard Educational Service.

Based on the successful series of filmstrips and slides produced by the Service, the new books are expected to become popular with both student and teacher.

"Principles of Electrostatics" is a 32-page book measuring 15cm. x 21cm. Its 13 sections each cover a particular aspect of electrostatics. Typical headings are: Insulators and conductors; the electrification theory; the gold leaf electroscope; capacitance and capacitors.

Minibooks are available from The Mullard Educational Service, Mullard Ltd., Mullard House, Torrington Place, London, W.C.1, price 2s. 6d. (including postage), cash with order.

GRUNDIG MANDELLO RESTYLED

Grundig (Great Britain) Ltd. announce that their popular Mandello Stereogram has been restyled to bring it up to date. For those who treasure their discs, the four speed de-luxe record changer incorporates a micro-lift facility. The smart push button radio receives v.h.f., m.w., l.w. and s.w. (band-spread) and can be fitted with a decoder when stereo broadcasts become available.

Two Grundig Superphon ceramic magnet loudspeakers provide 6W of stereo power—3W per channel—and sockets are provided for the connection of extension loudspeakers, stereo decoder, tape recorder and external aerials. The Mandello incorporates a dipole and ferrite aerial, 6 valves, metal rectifier and 2 diodes. Price is 99 guineas including P.T.

.. COMMENT

SILVER MEDAL AWARD FOR MULLARD FILM

The Mullard film, "Thin-film Microcircuits", was awarded a silver medal (1st prize) in its category at the 10th International Festival of Scientific-Teaching Films, organised by the University of Padua in conjunction with the 1965 Venice Film Festival. Over 150 films from 18 countries were entered and of these 52 were selected for showing to the international jury.

Britain did particularly well this year. Seven of the ten films entered were chosen for screening and three gained awards.

"Thin-film Microcircuits" is a 16mm sound and colour film which deals with the manufacture of this new type of electronic component from design stage to the finished product. The film also describes typical applications including space vehicles, miniature computers and industrial electronic equipment.

Another Mullard film, "Electromagnetic Waves—Part 2", won a bronze medal (2nd prize) at Padua last year.

Both of these films were featured at the P.W. and P.T.V. Filmshow on Feb. 4th (see page 1030).

ELECTRONICS GROUP

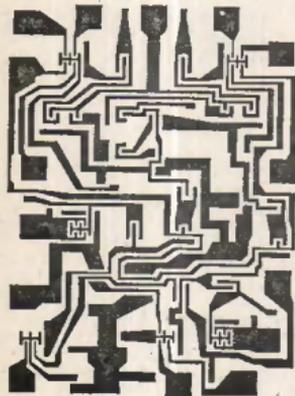
The St. Cyres Electronics Group are holding a series of meetings in Penarth (Glam.) the first of which was on 14th January, 1966.

By kind permission of Mullard Ltd., this group will be showing a very wide range of technical films which they have issued. As Penarth is only about 5 miles from Cardiff, many readers may be interested in joining the Group and seeing these films.

Further details may be obtained from the organiser, Mr. C. Bogod, "Dickens", 26 Forrest Road, Penarth, Glamorgan.

HAM COMPETITION

An international competition for radio hams is being held in Tenerife. The competition is based on the total number of contacts made during a given period, between operators all over the world and those in Tenerife. Allowance is made for the distances involved. Prizes include a 20-day trip to Tenerife for two people including accommodation in a first-class hotel. Diplomas and Silver Cups will also be awarded.



ELECTRIC MODERN ART

Resembling an artistic design of modernistic proportions, this pattern is actually an industrial creation made by electronic engineers at the molecular electronics division of the Westinghouse Corporation in the USA. The pattern is to be reduced thousands of times until it is finally engraved on a piece of silicon of almost microscopic size.

The microminiature will then become an integrated circuit for use in a computer.

Any Ideas Please?

CAN any of your readers suggest a practical design for a simple transistorised audio generator tuning over a range of about 20—200 c.p.s.?

This would be of use in tuning loudspeaker enclosures and would not require a critical waveform output.

F/Lt. G. Halbert.

Officers' Mess,
R.A.F. Lyneham,
Wiltshire.

Frequency Synthesisers

I WAS interested to see in the March issue a description by "Henry" of a piece of apparatus known as a Frequency Synthesiser. He appears, however, to have grasped the wrong end of the wrong stick. As I understand it, and I have used frequency synthesisers, the equipment has nothing to do with diversity reception or multi-channel transmission as Henry would appear to believe.

The frequency synthesiser is a device which combines the stability of a crystal oscillator with the flexibility of a v.f.o. It can give any frequency from 1kc/s to 30Mc/s in 1kc/s steps with an accuracy of ± 2 parts in 10^6 and is normally used by s.w. broadcast stations as the r.f. drive input to a transmitter.

The basis of the synthesiser is a 1kc/s crystal whose harmonics up to 30Mc/s are selected by a triple mixing system to give the required output frequency. It uses about 90 valves and would cost a wealthy ham about £1,500.

J. N. Douglas.

Glasgow, N.W.,
Scotland.

Sell or Loan

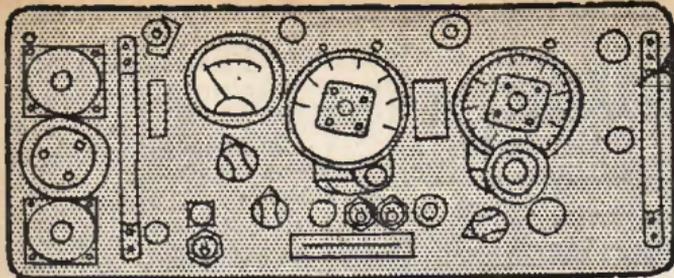
Sir, I would be grateful if any reader could sell or loan me...

...ferrite rod aerial for Cosor 524 a.m./f.m. receiver.—T. F. Jones, Flat 2, Block 3, Wychbury Court, Highfields Estate, Halesowen, Birmingham.

...a circuit diagram, and any information (i.e. modern equivalents of valves) etc. of the R1155A and R107 ex Government receivers.—Brian Carling, A4941, 27 Ellis Close, Cottenham, Cambs.

...the circuit diagram of the Sound A20/03858 tape recorder.—N. McFerran, 58 High Street, Newtownards, Co. Down, N. Ireland.

...the circuit diagram or service manual for the Air Ministry Oscilloscope type 11 Ref. No. 105/562.—R. E. Fields, 49 Torkington Road, Gatley, Cheshire.



No 19 SET MODS

by S. Simpson

Part 2

Inclusion of output stage

The reader should now have a highly sensitive receiver which, using about 10ft. of aerial, will provide signals all over the 4.5 to 8Mc/s dial. The 2 to 4.5Mc/s band will be relatively quiet at the 2Mc/s end until dusk. One does not always use a headset, however, so the next modification is the provision of loudspeaker output.

- (1) Strip all components and wiring from above and below the v.h.f. end of the chassis, other than the valve sockets and the input plugs. (If desired, the input plugs can eventually be removed and the holes blanked off but, at present, the h.t. and l.t. live leads must be retained for two more checks before the mains power supply is introduced. Take care not to damage the two-gang v.h.f. variable capacitors, to be used later.
- (2) Remove B GAIN and in its place, fit an s.p.s.t. toggle switch (hereafter "LS switch"). Mount the switch horizontally and ensure that ON entails moving the dolly towards the centre of the panel so that it "brings on" the output stage when wanted.
- (3) Check that V8A, pin 8 is earthed. Connect a heater lead from V8A, pin 7 to the ON

contact of the LS switch. Connect the OFF contact of the LS switch to V3A pin 2.

- (4) Make up the circuit for V8A as shown in Fig. 4. The screened, grid-input lead is connected to TS1 tag 4A (already carrying a phone output lead; the lead remains there). The screen is earthed at TS1, tag 3.
- (5) An output transformer was set aside earlier. This should now be fitted adjacent to V8A (Fig. 2). The blue terminal is connected to V8A, pin 3, green to TS1, tag 9, red to one lead of a 10in. length of twin cable, and white to the other lead; white is also earthed.
- (6) Remove the KEY socket; note the required contacts to adapt the KEY socket to serve as a non-self-shortening loudspeaker plug. Connect the twin output lead to these contacts, then refit the socket.
- (7) Fit one of the two 6V6 valves (supplied with W.S.19) to V8A socket. Connect the loudspeaker to the LS plug and insert the plug into the LS (formerly KEY) socket.
- (8) Set up the receiver for phone operation, and when assured signals are being received, set the LS switch to ON. The loudspeaker should operate after the usual warm-up time for V8A. Switch off LS and check the loudspeaker output fades out.

Inclusion of band-spread

The tuning drive on the four-gang capacitor is quite slow, but although satisfactory between 2.5 to about 3.5Mc/s, signals whip through the dial rather swiftly at higher frequencies. A very satisfactory band-spread can be incorporated using existing material, but entails a fair amount of dismantling of the P.A. tuning arrangement.

- (1) Dismantle the P.A. tuning drive as under:—
- (a) Remove the central screw on the FLICK/SET control; remove the knob.
- (b) Remove the four flick pre-set screws from the centre-piece on the P.A. TUNING dial.
- (c) Remove two 4BA screws securing the complete drive and capacitor assembly to the chassis. The assembly should now be loose, but not yet removable.
- (d) Remove one 4BA dome-head screw securing the slow-motion drive assembly to the panel.

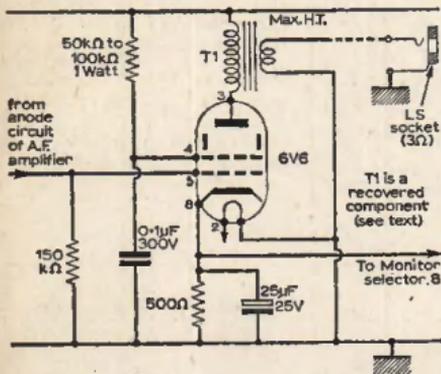


Fig. 4: Output stage, showing monitor point.

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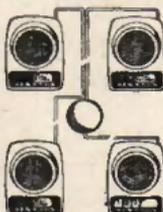
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(Note that there is a stout leaf spring holding the slow motion spindle against the edge of the engraved dial.) Remove the drive assembly.

- (e) Loosen the grub-screw in the side of the dial centre-piece; it should now be possible to separate the dial from the flick-stop assembly; this will expose a small tapered pin which passes through the collar of the flick-stop stub-flange and the capacitor shaft. This collar (not the pin) is required for the modification and the pin must therefore be removed. (Note that the pin is fairly soft and easily burred. One way of removing it is to cut away the tapered end with a hacksaw and punch out the remainder.)
- (f) Summing up the results: One should have available the slow motion drive and attaching 4BA dome-head screw, the engraved dial riveted to a metal disc to which the centre-piece is attached by one 4BA dome-head screw, the collar and stub-flange removed from the capacitor shaft. At this stage it is advisable to prepare a new dial to replace the existing engraved plate, which can be removed by withdrawing the 4BA screw from the centre-piece and drilling out the four small rivets securing the plate to the metal disc. In the author's version a circular disc of thin card and of the same diameter as the engraved plate was marked with 30 equidistant divisions around one semicircle and Letraset numerals 1, 5, 10, 15, 20, 25 and 30 attached at the relevant divisions. The centre of the card was removed to almost the diameter of the underside of the centre-piece, then the card disc was secured to the metal disc, using Cowgum. The centre-piece was then refitted to the metal disc to complete the dial assembly.
- (2) Earlier the reader removed the v.h.f. two-gang capacitor mounting plate. This assembly is now used as the bandspreader. Fit the collar to the capacitor shaft and insert the collar through the panel aperture formerly occupied by P.A. TUNING; leave the mounting plate as it stands for the moment.
- (3) Fit the dial to the collar (note the flat on the collar which mates with the dial). Fit the slow-motion drive to loosely mate with the dial.
- (4) By trial assess what shimming is necessary between the bandspreader mounting plate and the receiver chassis to obtain adequate slow-motion drive on the dial. Also check for positioning of the capacitor for all-round smooth drive. Mark the chassis, remove the capacitor assembly, drill and tap the chassis (there is no

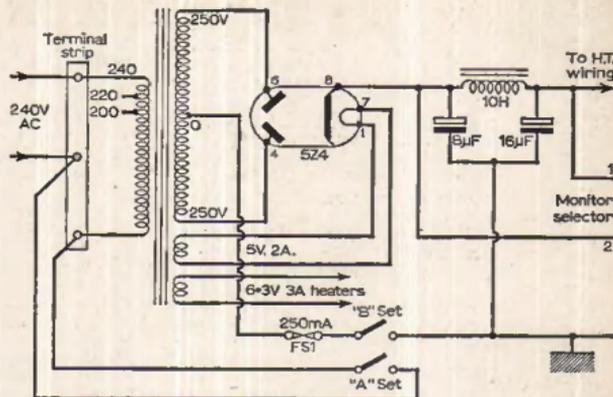


Fig. 5: Power supply circuit, showing monitor points and h.t. control.

access for nuts under the chassis), then fit the complete assembly.

- (5) Set the dial to 30 opposite the right-hand index mark. Fully enmesh the capacitor vanes, then lock the capacitor shaft, using the grub-screw accessible in the stub-flange.
- (6) Strap both sets of fixed vanes and leave approximately 10in. tail on the lead.
- (7) At the four-gang capacitor unsolder and remove the compression trimmer at OSC (marked on top of the capacitor plate). Pass the band-spreader tail through the trimmer hole, assess and cut the length to reach the rear connection, i.e. to the fixed vanes. Withdraw the lead, prepare and heavily tin about $\frac{1}{2}$ in. of the tail, reinsert the lead and solder it to the rear tag.
- (8) Set up and check the receiver to try out the bandspreader. Note that it has little effect around 2-3Mc/s but is extremely useful at higher frequencies, drawing "attention" to dozens of c.w. signals usually unnoticed without band-spreading.

Inclusion of Power Supply

The reader, running through this article as "something to read", may wonder why a mains power supply has been left so late; the reason lies in the weight of the W.S.19, even as found, and the lifting and turning necessary during the preceding changes. If a power supply were added as an early step the work of the succeeding steps would be quite laborious. To add the power supply proceed as follows:

- (1) Obtain a top-of-chassis mains transformer to supply about 250V at 50mA, 6.3V at 3-0A, 5V at 2A, also a smoothing choke giving about 10H at 50mA. A smoothing capacitor assembly providing 8µF and 16µF at 300V working, a 5Z4 rectifier and a three-way tag-board complete the components list.
- (2) The circuit diagram is standard and given in Fig. 5. The locations of the components are shown in Figs. 1 and 2.

- (3) The mains input lead terminates on the three-way tagboard (if possible attached to the choke) which will carry a flat-twin mains lead to the A SET switch.
- (4) Dismantle the two-way switch assembly to gain access to A SET and B SET switches.
- (5) Cut the existing lead on A SET passing into the W.S.19 power input plug cableform. Parallel-connect the freed contact on A SET and its neighbour at that end. Add one wire of the flat-twin mains lead (sub-paragraph 3) to the parallel connection.
- (6) Remove the l.t. connection at the other end of the switch, parallel the two tags at that end and add the remaining wire of the mains lead.
- (7) Ensure that all l.t. leads (less that passing into the W.S.19 power supply cableform) which were associated with the A SET l.t. connection are still in good contact. Fit a short length of sleeving over the common connection and tuck safely away.
- (8) At the B SET switch remove the lead passing into the W.S.19 power supply cableform, cut the lead short at the power input plug. On the B SET remove the lead supplying h.t. to the circuitry and connect it to the tail from the smoothing choke (see Fig. 5). Connect the contact on B SET, just cleared, to chassis. Connect the contact which carried the W.S.19 h.t. lead to the centre tap of the h.t. winding on the mains transformer. The B SET switch now provides means of disconnecting h.t. while the heaters remain active.
- (9) At the remote end of the mains supply cable check for continuity between leads with A SET "on" and open-circuit with A SET "off". Check for open-circuit to chassis with A SET in both positions.
- (10) Connect a voltmeter (to read 300V) between the smoothing choke output and chassis.
- (11) Ensure A SET is switched off and l.s. is also off. Connect the mains supply cable to a power outlet. Switch on A SET and B SET, and note that the valve heaters are alight.

Watch for the developing h.t. in the voltmeter. Switch off A SET.

- (12) Plug in the headset, turn up A GAIN and R.F. GAIN. Connect the aerial and switch on A SET. Check the receiver functions normally, then turn down A GAIN to minimum and check for hum on the h.t. supply. If all is well switch off; if hum is present make the usual checks of capacitors, leaking cathode insulation and so on.
- (13) Fuses. If it is desired to fit fuses a panel fuse to carry h.t. can be installed in the QUENCH aperture and another may be fitted in the rear wall of the chassis to carry mains input. The h.t. fuse should be connected in the earth return from the h.t. winding of the mains transformer (Fig. 5). Suitable fuses would be 250mA (h.t.) and 2A (mains).

Monitoring system

The inclusion of a monitoring system is a fairly simple task but, in a complex receiver, an addition well worth having. The system involves a nine-way single-pole wafer switch instead of the existing five-way double-pole component and uses the existing meter. The monitor system is shown in Fig. 6.

- (1) Remove the panel meter and clamp assembly held by one 4BA screw passing through an eyelet on the clamp.
- (2) Remove the clamp from the meter by loosening an 8BA screw at the top of the clamp.
- (3) Using a 10,000 Ω series resistor check the meter on a 1.5V supply for continuity. If satisfactory proceed to step 4 and then to step 7. If the meter is faulty continue as follows (these instructions are based on the author's findings on a faulty meter and the fault may be common in these meters, now several years old).
- (4) Withdraw three small screws from the side of the meter case, withdraw the meter from the case and ensure the needle is free to move.

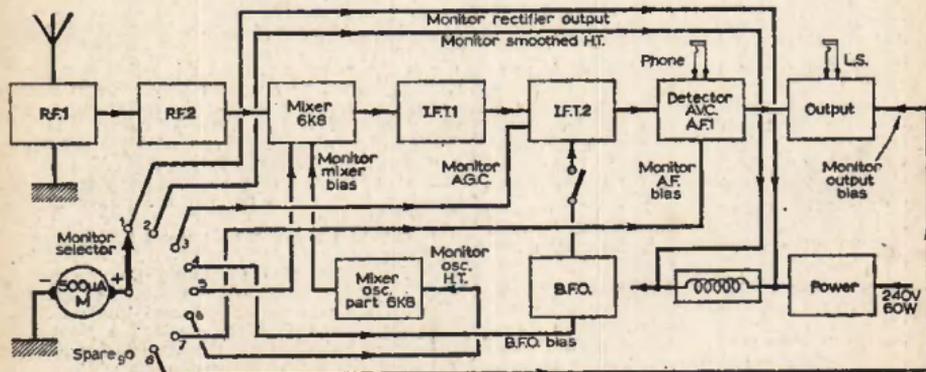
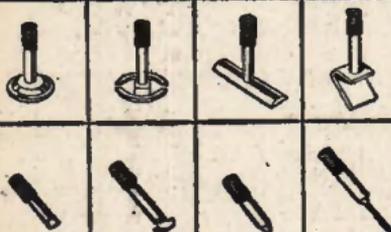


Fig. 6: Monitoring system in modified W.S.19 receiver.

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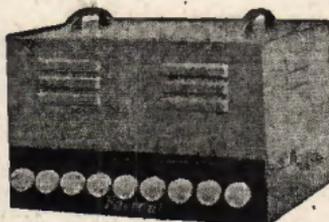
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- (5) Again check for continuity but this time across the movement lead-in connections, i.e. close to the hairspring. If no reading is obtained check, using a lens, for broken connections. If there are no breaks the coil is damaged beyond amateur repair.
- (6) If a reading is obtained when the movement is checked at the lead-in connections look for a minute break at the termination of a small coil on top of a resistor attached to one input terminal. (The break is probably caused by wire, hardened by age and subjected to fairly severe vibration during service life.) Resolder the connection and check the meter from the external terminals.
- (7) Remove the existing dial. Prepare a new dial numbered one to ten and attach it to the existing dial (as was done with the bandspreader dial). Refit the modified dial to the meter.
- (8) Reassemble the meter and put it aside meantime.
- (9) Cut all wiring to the monitor switch. Remove the switch.
- (10) Fit a nine-way single-pole switch in place of the removed monitor switch. (Space is limited, therefore the diameter of the new switch must not exceed 1½ in.)
- (11) Obtain about 15 in. (at least) of nine-way multicore cable, preferably having leads of different colours. Strip the outer sleeve (and screen if any), then prepare ½ in. tinned connections at one end of the cable. Pass the untreated tail of the cable through one of the two eyeletted holes adjacent to the bandspreader capacitor.
- (12) Turn the monitor switch knob fully counter-clockwise and locate contact 1, also the wiper contact. To contact 1 solder one tinned lead of the cable. Follow on with the remainder of the cableform, noting on a monitor schedule the colour of each lead and its relevant contact on the monitor switch.
- (13) When all leads have been attached to the switch contacts solder a 3 in. lead to the switch wiper.
- (14) Refit the meter. Earth the negative terminal. To the positive terminal connect the switch wiper lead.

The selection of monitored points is, perhaps, something for the reader to decide since views on this subject are many. The author suggests the following as being very useful (reasons are given against the less obvious points):

- (1) Main h.t. at rectifier cathode.
- (2) Main h.t. at choke output. If a fault changes the h.t. current demand noticeably this is shown by a change in the known difference between the cathode voltage and the correct value of choke output voltage.
- (3) Automatic volume control (measured by its effect on cathode bias of V1C. This setting provides visual tuning control and indication of signal amplitude but only on R.T.)
- (4) Beat-frequency oscillator cathode bias. (If no b.f.o. note can be obtained this setting provides a quick check on (a) the presence of h.t.

at the valve, (b) whether or not the valve is oscillating. In case (a) there is no bias reading, in case (b) the reading is above normal.)

- (5) Frequency-changer bias. (If the oscillator section of the 6K8 fails the bias reading is slightly, but noticeably, greater than when the oscillator is normal. Two conditions are noted in the monitor schedule: (a) Normal operation at 4.5Mc/s, (b) bias with the oscillator dead.)
- (6) Frequency-changer oscillator h.t.
- (7) First a.f. amplifier cathode bias.
- (8) Output stage cathode bias.
- (9) In the author's receiver this position is used to monitor the +28V supply to a teleprinter used in conjunction with the receiver.

All of the above are shown in Fig. 6. Whichever circuits are chosen the value of resistor to be connected between the monitor point and the cableform lead can be found from the Ohm's Law formula:

$$R = E/I \text{ where } R \text{ is in megohms, } E \text{ in volts, and } I \text{ is a fixed value, } 250 \text{ (this is the half-scale reading of the monitor meter).}$$

For example, supposing one wants to monitor a 10V cathode bias point,

$$R = 10/250 = 0.04M\Omega = 40k\Omega.$$

Since the readings need not necessarily be exactly half-scale any value between (say) 37k Ω and 44k Ω should be satisfactory.

Two points which should have the same resistor value, however, are the rectifier cathode h.t. and the choke output h.t., the reason being that one requires a comparison of meter readings to show the voltage drop across the resistor in the h.t. supply.

When all wiring has been completed in the monitoring system the cableform should be laced with button thread to tidy off. The receiver should then be run up and allowed ten minutes or so for temperature stabilisation, then the various readings taken down on the monitor schedule, which should be attached to the receiver, or filed for reference.

The inclusion of a monitoring system completes the series of modifications carried out by the author. If no snags have been encountered and the original receiver was in reasonable condition to begin with the reader should now possess a receiver whose performance in the 2 to 8Mc/s band is the equal of many very much more expensive receivers. If he has had experience in alignment of receivers he may be able to improve this high performance still further by slight co-ordinated adjustment of the four compression trimmers situated in the area to the rear of the waveband switch. (The fifth trimmer, near the panel, alters beat frequency and should be left alone.) If he has no realignment experience leave these four trimmers to someone with the necessary knowledge to adjust them.

Ensure all superfluous wiring and wire clippings are cleared out of the chassis and that new wiring, components, etc., will be clear of the inner base when fitted. Drill the outer case to pass the mains cable (and provide access to the mains fuse if fitted), then fit the receiver into its outer case to complete the job. ■

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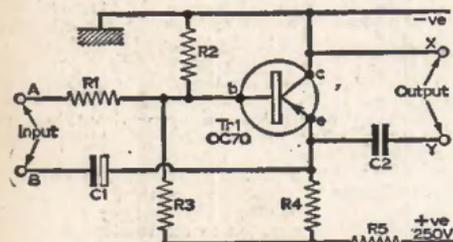


Fig. 1: The single-transistor circuit of the unit.

The writer has used the preamplifier as a second microphone input to an existing valve amplifier with excellent results. The output of the pre-amplifier was connected to the gramophone pick-up sockets; the gramophone volume control thus became the second microphone gain control. Being only 3in. x 1½in. x ¼in. in size, the unit fits easily into a tape recorder, care being taken to keep the transistor away from components which run hot, i.e. valves and smoothing resistors.

COMPONENTS LIST

R1 100 Ω	C1 100 μ F 12V
R2 100k Ω	C2 0.1 μ F 150V
R3 470k Ω	Tr1 Mullard OC70
R4 5-6k Ω	
R5 330k Ω	

All resistors ¼W 5% high-stability.

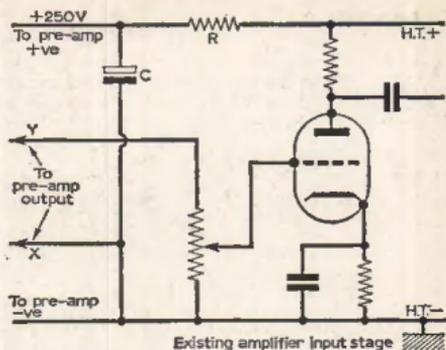


Fig. 2: The arrangement of a valve amplifier input stage for connection of the preamp. R has a value to regulate the amplifier h.t. to 250V for the preamp. C = 8 μ F 350V.

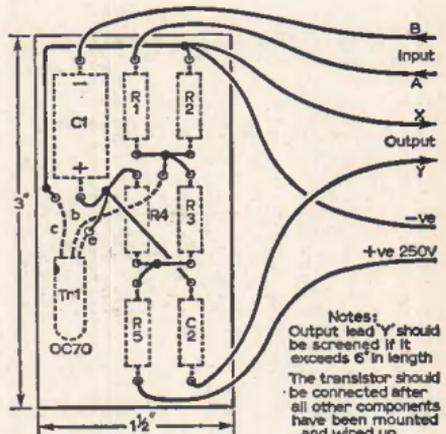


Fig. 3: Construction of the unit is on a piece of paxolin and the above layout was used successfully by the author.

Construction

The preamplifier can be built on a paxolin strip 3in. x 1½in. The strip should be drilled to take the wire ends of the components, which are twisted together behind the strip and soldered—shown by heavy dots in Fig. 3. Input, h.t. and output connections are made with flying leads.

Before connecting to the h.t. supply check the polarity of the preamplifier. Successful operation has been obtained with an h.t. supply of between 150-260V. Should the valve amplifier h.t. be in excess of 260V it can be reduced to a safe level as shown in Fig. 2. This also shows the method of coupling the preamplifier to the valve amplifier.

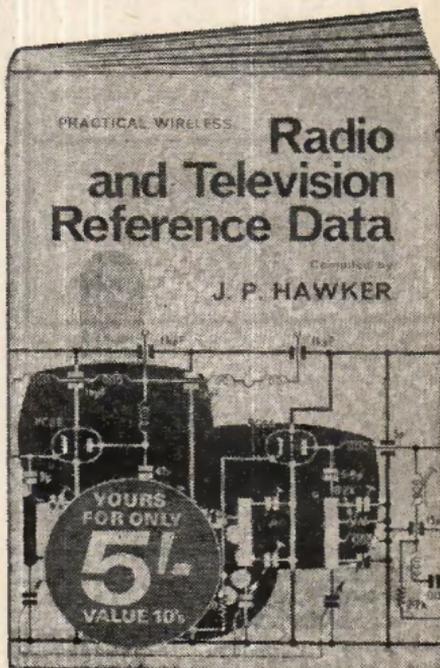
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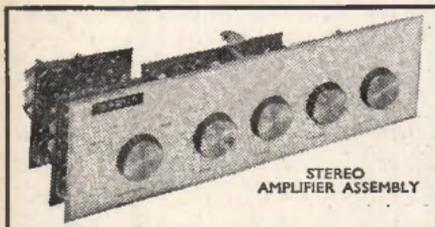
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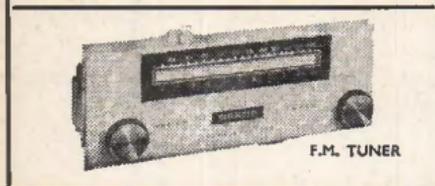
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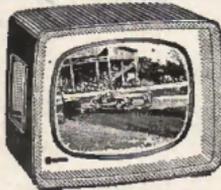
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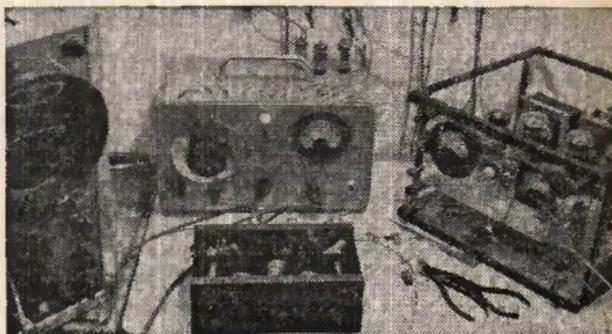
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An ELECTRONIC GATE OR TRACE DOUBLER



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H. T. Kitchen

THE majority of oscilloscopes used by the majority of amateurs have one characteristic in common, the inability of displaying two separate signals as two separate and distinct waveforms on their cathode ray tubes at the same time. They are, in other words, single beam oscilloscopes.

Now it is quite probable that many amateurs will never need, or want to, view two different signals simultaneously, whilst others, like myself, will only need to do so from time to time. Undoubtedly the easiest way of doing so is by using a double beam oscilloscope, though this type of instrument, due to its high cost, can almost certainly be regarded as an unjustifiable luxury in all but a few exceptional cases. We must not, however, overlook the amateur who has either bought or constructed a first-class single beam oscilloscope for, having spent a goodly portion of his hard-earned income upon it, he will not exactly be overjoyed at having to start again.

Fortunately it is not unduly difficult to obtain

double beam operation from a single beam oscilloscope, though not, unhappily, with the same facility as would be obtained from a genuine double beam instrument. This is brought about by the use of a special electronic gate into which the two signals are fed simultaneously. The gate then allows each signal to be passed turnabout to the oscilloscope, suppressing in the meantime the signal that is not being passed, the net result being two separate waveforms corresponding to the two separate signals.

Gate Principles

The principle of the electronic gate is quite simple and is illustrated in Fig. 1. It comprises two amplifiers for the Y1 and Y2 signals and a gate generator which provides two gating signals, these being square waves of identical frequency but out of phase by 180°. These are also fed into the Y1 and Y2 amplifiers and alternately allow them to conduct and then drive them into cut-off.

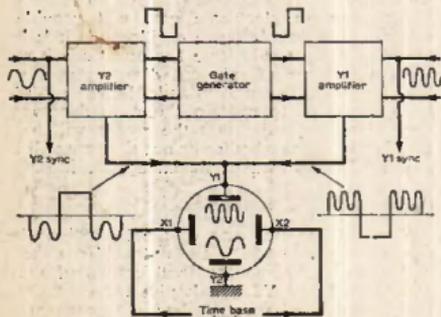


Fig. 1: Diagram showing operation of the electronic gate.

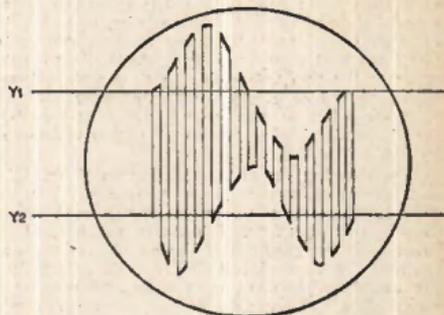


Fig. 2: The effect of a slow gating speed. The Y1-Y2 waveforms are composed of positive and negative peaks of the gating waveform.

Being in anti-phase it follows that when the Y1 amplifier is conducting, the Y2 will be cut off and vice versa, so that the c.r.t. is presented with a sequence of "samples" which, if the gating waveform is of the correct frequency, gives the impression of two separate traces. If the oscilloscope timebase is adjusted to display the gating waveform the Y1 and Y2 signals will be seen to occupy the horizontal portions of the square waves' positive and negative peaks. This effect, which is shown in Fig. 1 and also in Fig. 2, is only shown to explain how the gating waveform provides two separate traces. In actual use the timebase will be used to display the input signals only and the gating waveform will not be visible except under certain circumstances. These will occur when a fairly slow gating speed is used and the input signals will be seen to consist of a series of dots or dashes, depending on actual frequency, which are really the positive and negative peaks of the gating waveform. At high gating speeds it is possible for the intervening spaces to be rendered luminous by the gating waveform and occasionally for the waveform to be faintly discernible. With practice it is possible to ignore both effects and concentrate only on the waveforms to be examined.

With a double beam oscilloscope each trace has its own vertical shift control and so can be positioned just where it is wanted. The two traces can be superimposed, transposed or one removed right off the screen. With an electronic gate, however, the two traces will move as one every time the oscilloscope's vertical shift control is operated. The superimposition and transposition facilities can, however, be retained, though neither trace can be removed off the screen without going to some, and usually unnecessary, trouble anyway. The gating waveform, as already explained, is a square wave which is fed into the two amplifiers out of phase by 180°. If the gains of the two amplifiers are exactly equal the two anti-phase signals will cancel out and in the absence of any Y1 or Y2 signals will show on the oscilloscope as a straight line. Where Y1 and Y2 signals are present they will be superimposed, occupying, as they do, a common baseline. Although this may suffice for some applications it is an unsatisfactory way of doing things, for there is every possibility of the superimposed traces causing misunderstandings and mistakes. It is desirable to either separate the two traces so that they are no longer superimposed or else to make one much brighter than the other or even to make the best of both worlds by combining them and neither objective is particularly difficult to achieve.

It was earlier stated that if the two Y amplifiers had equal gains the gating square wave would cancel out and the oscilloscope would simply display a straight line. It must be understood that since both the gating and the external signals are fed into the Y amplifiers this remark will apply to the gating waveform's gain only and not to the signal gain. If the gain of the two amplifiers to the gating waveform is made unequal the output will be a square wave and it will be possible to separate the two external signals. This is done most easily and conveniently by varying the bias

SPECIFICATION

Power Supply Required:

250V at 20mA
6.3V C.T. at 1.2A

Y Amplifiers:

Freq. Response: 15c/s—50kc/s ± 1 dB
10c/s—200kc/s ± 3 dB

Input Impedance: 1M Ω

Gain: x5

Gate Generator:

Fixed frequencies of 50c/s, 175c/s, 500c/s, 1750c/s, 5000c/s approx.

Gate Output: 85V P-P (anodes of V1)

Gating Transients: 11V P-P (cathodes of V2 V2)

Note 1

All static voltages except multi: V1 measured with VRI at mid position and with S1 at "off" position using 250V and 10V ranges of 20k Ω /V meter.

Note 2

Gain and frequency response of Y1 Y2 amps measured separately with VRI at mid position and gate at "off".

on the gating stages, so causing unequal anode currents to flow in the common anode load resistor, thereby providing the required unequal outputs. It is customary to label the top trace the Y1 and the lower the Y2 trace and to label the input sockets accordingly, thereby making it possible to identify the two signals.

It is also possible to make one trace more or less brilliant than the other by varying the mark space ratio of the gating square wave. The two traces will be equally brilliant when the gating waveform has a mark space ratio of one to one. If the mark space ratio is altered the two traces will differ in brilliance since one waveform, the more brilliant of the two, will be allowed to remain longer on the cathode ray tube than the dimmer trace which will have been allowed the least time. In practice the mark space ratio can be altered by providing the multivibrator with unequal grid leak resistors or by making the cross-coupling capacitors unequal in value. It is not usually possible to alter the mark space ratio to any great extent without adversely affecting the waveform.

Although an electronic gate will provide the services of a double beam oscilloscope it is by no means a perfect substitute for one since it suffers from a number of inherent disadvantages which limit its scope (no pun intended!) and usefulness. It is proposed to explain these disadvantages in some detail in order that the reader will be in a position to judge whether or not the gate will be of any use for his own particular applications.

Limitations

Strangely enough one of the limitations is the oscilloscope into which the output from the gate will be fed. This output will consist of the two external Y1 and Y2 signals plus the gating square wave which has been described, with perhaps some justification, as the carrier wave. Now one of the properties of a square wave is its ability (unwanted

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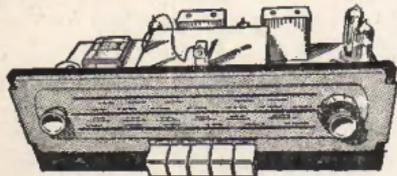
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in this particular application) to reveal the characteristics of the equipment through which it has been passed. Sagging or sloping tops will denote a phase advance due to a poor low-frequency response, whereas rounded corners will denote a phase lag caused by a poor high-frequency response. In really bad cases the emergent waveform will bear very little resemblance to the ingoing one. It will therefore be evident that for successful operation of the gate the oscilloscope's Y amplifier must pass the gating square wave without distorting it, at least to any excessive extent. If it is unable to do so the output from the gate will have to be fed directly to the cathode ray tube's Y1 plate, in which case some alteration to the deflection circuitry will probably be necessary. Where the oscilloscope's vertical amplifier uses symmetrical or push-pull deflection it will be necessary to earth the Y2 plate (capacitively if there is any z.f. voltage on it) and to feed the gate to the Y1 plate only. Due to the variety of deflection circuits in common use more explicit directions cannot be given, though it can be said that additional amplification before and after the gate will probably be necessary and will be due to the very small gain afforded by the Y1 and Y2 amplifiers plus their inability to deliver the very high peak-to-peak voltages necessary to fill the working area of the c.r.t.. It must be emphasised that these modifications will only be necessary with very simple oscilloscopes with restricted vertical bandwidths and it is quite possible for the gate to work, at least reasonably well, with the majority of amateur oscilloscopes. It is safe to say that if the oscilloscope can reproduce the gating square wave with reasonably vertical sides, square shoulders and reasonably horizontal positive and negative

peaks that it will work well with the gate.

Yet another limitation is the gating speed. In theory it should be half the timebase speed so that each waveform can be traced out in turn. In practice surprisingly good results can be obtained with very widely varying gating/timebase speeds. Some authorities recommend a low gating speed, others a high speed. My own experience is that very slow gating speeds cause the waveform to sparkle in a rather fascinating way which causes eyestrain rather more rapidly, whilst very high gating speeds cause the space between the Y1 and Y2 waveforms to be rendered luminous. Of the two the latter effect is preferable to the slow speed sparkling, but since individual preferences vary and since it is desirable to work at half the timebase speed whenever possible the gating waveform is made variable in frequency.

Bandwidth

The last disadvantage to be covered is that of bandwidth. By the very nature of its operation the bandwidth that can be covered is restricted to the audio and low radio frequencies and is due to the inability of the simple gating circuit used to generate the ideal square wave necessary to bring about satisfactory gating at high frequencies. It will nevertheless satisfactorily handle 200kc/s since waves as shown in the waveform diagrams for which purpose the two inputs were temporarily connected in parallel. The distortion evident was present in the input signal which was derived from a transistorised "Nombrex" signal generator. This shows that in spite of the remarks concerning the recommended gating to timebase speed ratio of 2:1 (or 1:2) acceptable results can be obtained

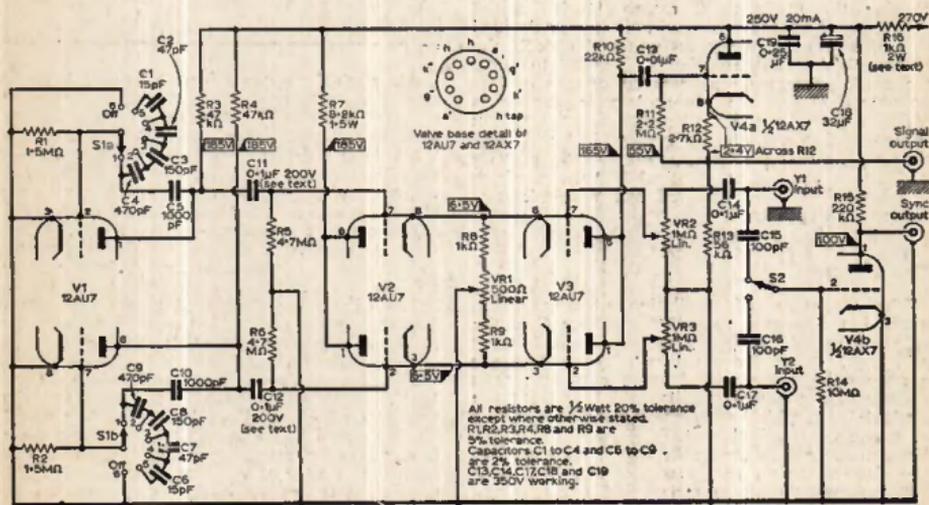


Fig. 3: Complete circuit of the trace doubler.

with widely differing ratios, for the gating speed used in this case was 1,750c/s. Although my own tests ceased at 200kc/s it is not unreasonable to suppose that even higher frequencies could be handled, though the signal gain would be well down when compared to the reference frequency of 1k/c. These remarks apply to sinewaves, the square wave response deteriorating after about 25kc/s, after which considerable rounding off occurs. The waveform diagram shows a square wave with a frequency of 25kc/s as generated Y1 trace and as reproduced at 1W output by a 10W amplifier on the Y2 trace. The rise time of the square wave was 2-0 μ S (the oscilloscope shows this as absolutely square, no rounding off, overshoot or sag) and as the Y1 trace is still reasonably square the performance of the gate as a whole can be considered to be eminently satisfactory—for all but the most exacting requirements anyway.

The complete circuit of Fig. 3 will be seen to be quite simple when split into its individual stages, which will be considered in turn commencing with the gate generator V1. This is connected as a cross-coupled multivibrator, the traditional form of square wave generator. Although the cathode coupled multi is somewhat simpler and cheaper to build the necessity of providing two anti-phase outputs precluded its use in favour of the cross-coupled variety. The variation in the gating speeds is brought about by S1a and S1b, which selects two sets of five close-tolerance (2%) frequency determining capacitors, C1 to C5 and C6 to C10. Two methods of discriminating between the two Y1 and Y2 traces have already been considered. It will be remembered that one method was offsetting the two traces so that they lay one above the other and the other method was making one trace brighter than the other by altering the mark space ratio. During initial development of the gating circuit it was decided that offsetting the two traces was undoubtedly the best method and it was thereby decided to make the mark space ratio as near to one to one as possible. Hence the close-tolerance capacitors which also confer another benefit upon the circuit. In conjunction with close tolerance (5%) grid leaks they ensure that no two positions of S1 provide oscillations at or near identical frequency, which could possibly have occurred had wide-tolerance components been used.

Time Constants

The frequency of oscillation of a multi is approximately equal to $0.77/CR$ where C is the value of one of the coupling capacitors and R the grid leak. Since large values of capacitors are correspondingly larger and more expensive than smaller ones the two grid leaks R1 and R2 are made 1.5M Ω , which means that the largest value of capacitor required is 1,000pF, for $f_0=50c/s$. For convenience in wiring the capacitors are wired in series so that the next capacitor value is 300pF (1,000pF and 470pF in series) and $f_0=1,700c/s$. The frequency therefore changes in five steps of 3-1 and makes it possible to select a gating speed most appropriate to the timebase speed within quite wide limits.

COMPONENTS LIST

Resistors:

All 20% 1W except where stated.	
R1 1.5M Ω 5%	R9 1k Ω 5%
R2 1.5M Ω 5%	R10 22k Ω
R3 47k Ω 5%	R11 2.2M Ω
R4 47k Ω 5%	R12 2.7k Ω
R5 4.7M Ω	R13 55k Ω
R6 4.7M Ω	R14 10M Ω
R7 8.2k Ω 1.5W	R15 220k Ω
R8 1k Ω 5%	R16 1k Ω 2W see text

Potentiometers:

VR1 500 Ω lin.	VR2 1M Ω lin.
VR3 1M Ω lin.	

Capacitors:

2% silver mica except where stated.	
C1 15pF 2%	C6 15pF 2%
C2 47pF 2%	C7 47pF 2%
C3 150pF 2%	C8 150pF 2%
C4 470pF 2%	C9 470pF 2%
C5 1000pF ceramic	C10 1000pF ceramic
C11 0.1 μ F 200V paper. See text.	
C12 0.1 μ F 200V paper. See text	
C13 0.01 μ F 350V paper	
C14 0.1 μ F 350V paper	
C15 100pF ceramic	
C16 100pF ceramic	
C17 0.1 μ F 350V paper	
C18 32 μ F 350V electrolytic	
C19 0.25 μ F 350V paper	

Valves:

V1, V2, V3 12AU7	V4 12AX7
------------------	----------

Valveholders:

4 89A

Switches:

S1 2 pole, 6 way, rotary
S2 1 pole, 2 way, miniature slide

Sockets:

4 co-ax.

Although the outputs from V1 could have been applied directly to the Y1 and Y2 amplifiers' V3, it was considered desirable to incorporate a cathode follower buffer stage V2 between V1 and V3. The gating waveforms are therefore developed across R7, VR1, R8, the cathode loads which are also common to V3, and V3 is therefore allowed to conduct and driven into cut-off, the two halves working in anti-phase. VR1 varies the bias voltages, increasing one as the other is decreased, so bringing about trace separation in the manner already described. The outputs from the two halves of V3 are developed across the common load resistor R9 and are fed via C13 to the cathode follower output stage V4a. It had originally been intended to use direct coupling until the voltage on the anodes of V3 was measured and found to be 165V. This would have necessitated a cathode voltage on V4a of about 170V, which it was felt was liable to put somewhat of a strain on the heater-to-cathode insulation of V4a. The output from the cathode of V4a was connected directly to the output socket without the customary coupling capacitor simply because the oscilloscope contains its own z.f. isolating capacitor and it was felt that two such capacitors in series were quite unnecessary.

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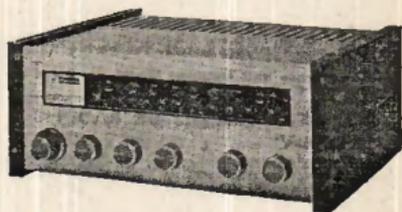
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TRACE DOUBLER

—continued from page 1078

V4b is arranged as the sync amplifying and limiting stage, more as a matter of wiring convenience than electrical necessity. The provision of a suitable sync signal is rather more involved than would be apparent at first. During the gate's early stages trouble was experienced with synchronising the trace using the internal sync position of the oscilloscope. A certain amount of jitter plus a reluctance of the trace to stay synchronised was finally put down to the gating waveform, the frequency stability of which is not

particularly good. The sharp leading edge of its square wave, however, synchronised the oscilloscope's timebase much better than a sine wave input having a much higher frequency stability, with the result that the sync tended to "wander" with the gating waveform. Once this was realised the sync signal was extracted before being gated and was fed into the oscilloscope, switched for external sync, with the result that jitter was absent and the timebase stayed synchronised to the input signal. S2 allows either the Y1 or the Y2 signal to synchronise the timebase.

CONTINUED NEXT MONTH

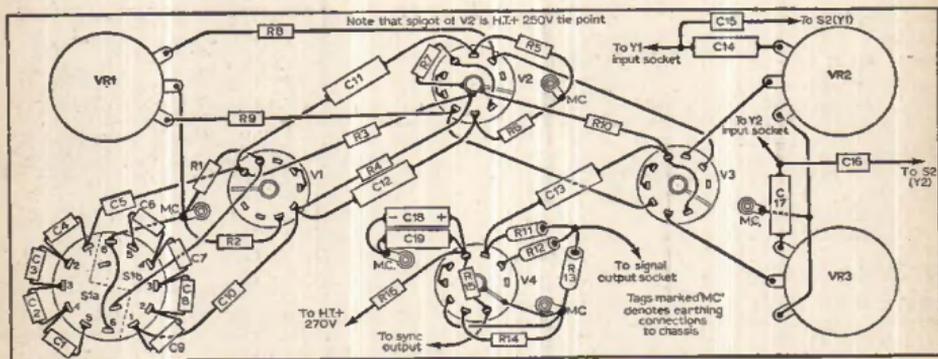


Fig. 4: Chassis wiring. The centre spigot of V2 is used as an h.t.+ point.

Reflex 2

—continued from page 1059

twists on this coupling were sufficient to put the r.f. amplifier in oscillation over the entire band. It was found that just a single twist is sufficient for most purposes, depending upon the nature of Tr1.

Capacitor C1 in the tuned circuit of L1 constitutes a form of padding capacitor, as well as serving as an r.f. bypass. Thus, the tracking can be influenced by varying the value of this component. Unfortunately, there is not available a miniature preset capacitor of sufficiently high value to permit easy adjustment here. The author found that an 0.005 μ F fixed capacitor satisfied the tracking for his reception area, but values above or below this could be tried since C1 is not difficult to change.

It is as well to remember that a simple set of this kind employing a sub-miniature ferrite rod aerial and only two transistors cannot be claimed to have super-sensitivity! Nevertheless, the prototype gave good volume on local stations, while after dark the more powerful European stations were received.

Housing the Receiver

The complete receiver is fitted in a small ready drilled plastic case as shown in the photograph which comes with back, tuning knob and paper

dial. The gang retaining nut being used to lock the assembly to the front panel and a hole in the side of the box to cater for the jack socket which, again, is held secure by its locking nut. The long 6BA bolt on the rear of the tuning gang is used to hold the back of the case in position.

The battery will have many months of useful life provided the carpiece jack plug is removed from the socket when the set is not in use.

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WHEN average transistors are used at low voltages two transistors will probably be necessary as a minimum in a resistance-capacitance oscillator. An advantage of the two amplifying stages as compared with a single transistor is that positive feedback can be obtained directly without resorting to the 180° phase shift of a ladder network. Simpler forms of network can therefore be used which have zero phase shift at the frequency of oscillation.

Wien Network

The most widely used of these is the Wien network (Fig. 3a), which has several less familiar variations, including current duals. The output from these networks reaches a peak equal to one-third of the input at the frequency where the phase shift becomes zero. With the surplus of amplification usually available the oscillator can be stabilised by an appreciable amount of negative feedback applied to reduce the loop gain to unity.

On subtracting the output of a Wien or similar network from its input to give the inverse response, i.e. by taking the new output from the other part of the network, a valley type response is obtained with a minimum equal to two-thirds of the input at the frequency of zero phase shift. In a true null network the minimum output would, of course, fall to zero.

This form of network, e.g. a bridged-T (Fig. 4a), can be utilised in a negative feedback path, but a large amount of positive feedback is required in an auxiliary positive feedback loop to bring about oscillation, so it is quite a difficult type of network to use successfully in the simpler types of oscillator. Since the net feedback will be positive, conditions approximate to those in a Wien bridge oscillator because when the input is a blend of positive and negative feedback it can be influenced by a phase shift in either.

Parallel-T

The true null network is a parallel-T. Although it has a rather different form of phase-shifting characteristic it can be substituted for the bridged-T network and, since the negative feedback now disappears altogether at the frequency of oscillation, considerably less positive feedback is required in the auxiliary loop.

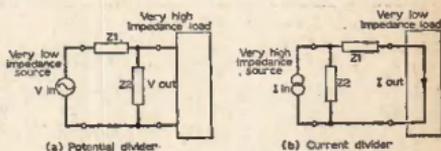


Fig. 2: Single stage networks.

The Parallel-T can likewise be rearranged to give an inverse response, which is a peak at zero phase shift, equal to the input and therefore three times that of the Wien network. Decreasing the impedance of the low-impedance branches will increase the output slightly above the input and it can then be used in an emitter follower type of oscillator where the actual voltage gain of the amplifier is less than unity (Fig. 7).

The disadvantage of the Parallel-T network is that it requires three capacitors and three resistors. The values of these are not all alike and require, for the null circuit, to be in a certain ratio, usually 2:1:1 as in Fig. 4b.

Tuning

In variable frequency oscillators tuning is easily accomplished by means of resistance capacitance networks. With their wide tuning range it is possible to cover from a few cycles per second to

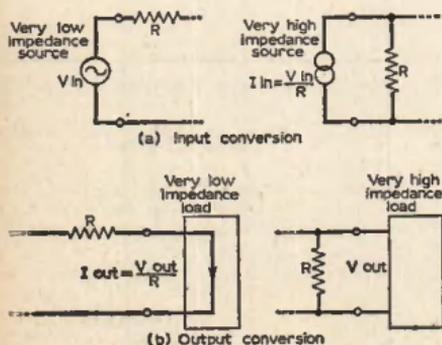


Fig. 1: Alternative arrangements at end of network where there is a resistance.

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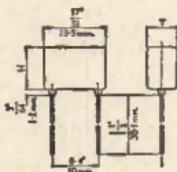
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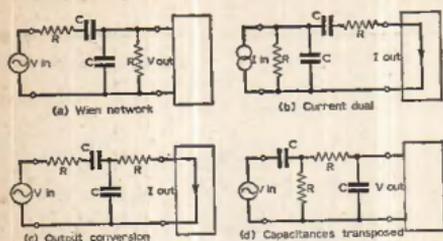


Fig. 3: Variations on the Wien network.

the lower radio frequencies in four ranges of switched capacitances, usually with a 10:1 ratio between adjacent ranges, and ganged potentiometers employed for fine tuning provide scales of about 300°. The potentiometers are normally wire wound to an inverse semi-log characteristic and of adequate resolution to enable a large-scale and sufficient accuracy of setting to be obtained.

Networks of the Wien type require only two capacitors to be switched per range and a dual-ganged potentiometer, by varying both of the network resistances simultaneously, avoids changes in the attenuation of the network, although altering the impedance it presents to the amplifying section. The amplifier normally requires to have a very low output impedance (and high input impedance), assuming a voltage transfer network, to avoid variation of the amplitude with tuning. A thermistor will probably be included to compensate for amplitude variations but it is advisable not to delegate too much of the task to the thermistor.

Amplifier Phase Shift

As far as possible all the frequency dependent phase shift should be concentrated in the R-C network, where it is determined entirely by passive components. The amplifier should be constant in performance and independent of frequency.

At high frequencies internal phase shift in the transistors can be minimised by employing r.f. transistors. At low frequencies amplifier phase shift can be overcome by means of direct coupling. A positive feedback loop nevertheless must normally contain a blocking capacitor to prevent it from exerting an unstabilising effect on the quiescent points of the transistors, but this may be one of the capacitors of the phase-shifting network. A negative feedback loop, on the other hand, has a stabilising effect on the d.c. levels and so may with advantage be direct coupled throughout.

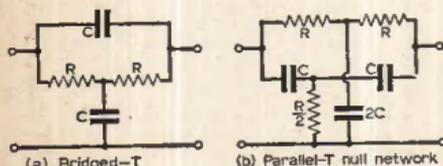


Fig. 4: Networks with valley-type responses.

Impedance Levels

Considering the frequency response as concentrated in the network assumes that the amplifier and network act independently. At both ends of the network, however, variable loading effects, if present, can interfere with the frequency response, making it less predictable and constant. Ideally the network must neither load or be loaded by the amplifier and instead of an impedance match for maximum power transfer the impedance levels of the amplifier and network should be widely different. This is necessary also to avoid distortion due to the non-linear input impedance of a transistor.

There are two ways of minimising the effect of a load variation. One is to make the source impedance much lower than that of the load, in which case the voltage remains nearly constant. The other method is to make the source impedance much higher than the load, in which case the current is nearly constant. When the source impedance is zero it is called a voltage generator and when the source impedance is infinite it is called a current generator. Each is represented by a special symbol in diagrams as in Fig. 1. In practice it is a question of the ratio between source and load impedances and we can speak of a "voltage input" when negligible voltage is developed across the load.

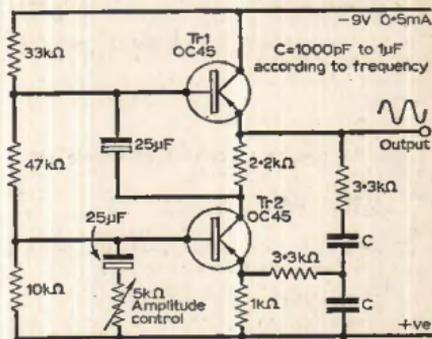


Fig. 5: Oscillator with modified Wien network.

A voltage transfer network, which subjects a voltage to its frequency response, can only be used when an amplifier has a low output impedance and a high input impedance relative to the impedance level of the network. For a current transfer network the reverse conditions apply. High input impedances and low output impedances in the amplifier can be secured by the use of emitter-follower stages. A special arrangement may be necessary for the bias supply to the base of the input emitter-follower, since a d.c. path must be maintained via a resistance to every electrode of a transistor. The high amplifier output impedance that is required for a current transfer network can only be provided easily when one of the resistors of the phase-shifting network serves also as a collector load resistor.

Network Duals

Voltage transfer networks can be regarded as composed of various potential dividers containing resistances and capacitances. Since two impedances in parallel constitute a current divider (Fig. 2b) it is possible to build current transfer networks on close analogy with the voltage transfer networks. To every voltage transfer network there corresponds a current transfer network called its current dual and it can be derived, for all the three-terminal networks we are considering, simply by turning the network round so that the input and output change places. The amplifier impedance levels must also be interchanged.

It is shown in Fig. 1 that a voltage generator in parallel with the resistance is equivalent to a current generator in parallel with the resistance. This is less obvious than the corresponding output conversion based on the current through a resistance being proportional to and in phase with the voltage across it.

It is convenient to be able to alter the feed arrangements of a network in this way but if the change is made only at one end the network maintains its frequency response but now derives an output voltage from an input current, or vice versa, so that the amplifier impedances can be both higher or both lower, as the case may be, than the network impedance.

A similar conversion can be made with capacitances at the ends of the Wien network if the change is made at each end so that the phase shifts cancel. By applying all of the preceding transformations to the Wien network quite a number of other networks with a similar frequency response can be derived. Some of these are shown in Fig. 3.

Networks appear to consist of simple sections but the loading of later sections upon earlier ones makes analysis more complicated. Separation of the sections of some networks is possible by inserting isolating stages between them, but this would be unusual in a transistor oscillator since it would increase the number of transistors.

Oscillator Circuits

It is possible to overrate current duals as oscillator networks and the typical a.f. oscillator uses a Wien network, which is of the voltage

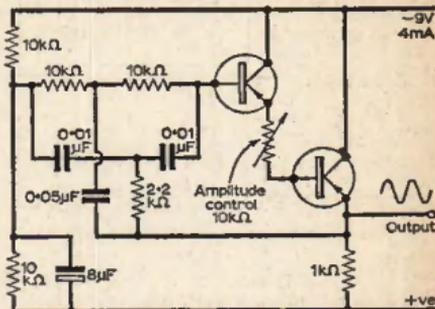


Fig. 7: Parallel-T +ve feedback oscillator.

transfer type. This requires a high input impedance in the amplifying section and with transistors this can be obtained from a compound emitter-follower arrangement involving two transistors. A third transistor will be necessary to secure a low output impedance and the amplifier must provide voltage amplification without phase reversal. The best direct-coupled arrangement combines p-n-p transistors with n-p-n. Increasing the number of transistors in an oscillator allows greater scope in design, enables the performance to be improved and permits variable tuning.

In a two transistor oscillator the possibilities are more limited, especially with direct coupling, and with less orthodox circuits snags are encountered such as the provision of satisfactory forms of amplitude control to prevent limiting.

The oscillator shown in Fig. 5, however, has a good performance and only consumes 0.5mA of current. It contains a Wien network supplied from the low-output impedance of the upper transistor, an emitter-follower, and from the Wien network an output current enters the emitter of the lower transistor, a ground-base stage, whose collector load resistance is increased in value artificially by the bootstrap action of the upper transistor. The current gain of the emitter-follower depends upon a sufficiently low impedance in its output circuit, hence the low resistance values in the network, but the equal capacitors C can be given a range of values to yield various spot frequencies. Amplitude control, gradual in action, is by varying the impedance in the base circuit.

The oscillator of Fig. 6 employs a Parallel-T null network in a negative feedback path. A Bridged-T would not give such a large undistorted output. Positive feedback between the emitters is varied for amplitude control.

An even better oscillator using a Parallel-T is shown in Fig. 7. In this optimum amplifier conditions can be secured with very high input impedance and a low output impedance. The low-impedance arms of the network are made lower in impedance than for a null network in order to step up the output voltage to the base of the first transistor. Amplitude control is by a resistance between the transistors but the low-resistance branch of the network could be varied instead. ■

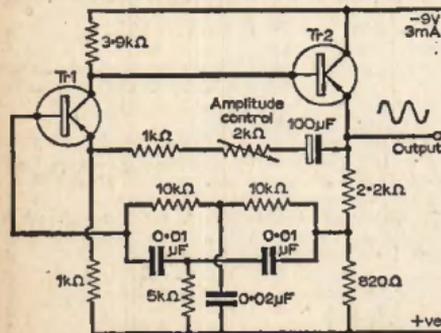


Fig. 6: Parallel-T -ve feedback oscillator.

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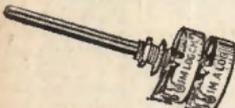
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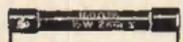
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CITY OF BELFAST Y.M.C.A. RADIO CLUB

ONE of the international Y.M.C.A.'s most senior organisations, City of Belfast Y.M.C.A. Radio Club (G16YM), is among the oldest radio clubs in the British Isles. The Club was founded in 1922 and received its transmitting licence and call sign in 1927, but there is some evidence of previous activity dating back to shortly after the close of World War I. It may be interesting to note that the Club licence holder, Mr. Robert Barr (G15UR), a "call" well known to the world's c.w. DX fraternity, has been continuously in membership since those early days and is still one of 6YM's most active members.

The lifetime of G16YM spans the history of radio experimentation in the north-east corner of Ireland and thus the City of Belfast Y.M.C.A. Radio Club can rightly be considered as the "cradle" of amateur radio in the most populous region of the six Irish counties comprising Northern Ireland. Certainly the majority of radio men and enthusiasts in the Belfast area, many of whom have made their mark on the "ham" hobby or in the wider radio world outside as Government scientific officers, Post Office, BBC and service technicians and operators, have all been connected with G16YM at one time or another.

In the Y.M.C.A. tradition of "Service, Not Self" the City Club membership takes modest

pride in its record during World War II days and was one of the few clubs which "carried on" in spite of cessation of transmitting activity and the loss of many of its most active members—of all ages—to the Royal and Merchant navies, Army and R.A.F. or engaged in security duties at home. Through its connection with the parent Y.M.C.A.—"Mecca" for thousands of Service men passing through Belfast—the City Club kept "open house" for Service "hams" from all over Ireland, Great Britain, the Commonwealth and the U.S.A. Many of these men held club office while stationed in Northern Ireland and needless to say, the war-time visitors' book is one of the most treasured items in the club archives.

Club accommodation is second to none and comprises a general clubroom on the third floor of the main city Y.M.C.A. building with a transmitting room on the floor above. The clubroom, recently refurbished in contemporary "decor", contains comfortable seating (including settee and easy chairs!) library with technical and current British and U.S. radio publications, listeners' corner complete with junior HRO, Morse practice equipment transcription record player and hi-fi amplifier for the club's audio section.

The well-appointed transmitting room on the floor above contains most of the equipment



As a YMCA organisation G16YM's emphasis has always been on youth. Pictured are some of the Club's SWL members and "junior ops." On the extreme left is G13UKS, Dave Vizard, the area's youngest licensee.



Left: Making checks on the Club's new S.S.B. transmitter are (from left) R. Barr (G1SUR), SWL Evans, F. Robb (G16TK) and Major Ian Kyle, club member and local T.A. "boffin".

Below: Tuning-up the standby T1154 transmitter are (from left) T. J. Moss (G13UFH) Club treasurer, J. Beattie (G13NOH) Club chairman, C. Rourke (G13IV) Club secretary, licence-holder R. Barr (G1SUR) and SWL member.

necessary for efficient operation under present-day conditions (the club recently made DXCC in spite of restricted hours of operation), the gear comprising a KW77 receiver, medium-power c.w. transmitter for 80 40 and 20m, 4m transmitter/receiver and all ancillary equipment. S.S.B. operation is also engaged in using a Sphinx transmitter, until the construction of the Club's own all-band S.S.B. rig is completed.

Main aerial requirements are taken care of by the recent construction and erection of a fine tri-band quad and supporting tower which has been mounted on one of the Y.M.C.A. building's highest points overlooking busy Wellington Place and already a city landmark. This erection also carries the v.h.f. aerial and will eventually be remotely controlled from the transmitting position. The emergency aerial is an all-band trap dipole.

The City of Belfast Club takes a full part in all local, national and international "ham" radio activity, chief among these being National Field Day and the popular Jamboree-on-the-Air, both events being held in conjunction with City Y.M.C.A.'s own Scout troop, which has a splendid QTH on the south side of the city.

Regular meeting nights are held each Wednesday and Saturday evenings from 8 o'clock but activities such as Morse practise and construction take place on other evenings as required and access to the club is available at all times. Any enthusiast—transmitting or SWL—whether "local" or just passing through can be assured of a real Irish Cead Mile Failte in the best "YM" tradition. Club secretary is Mr. Cedric Rourke, G13IVJ, 32 Kirkliston Park, North Road, Belfast 5.



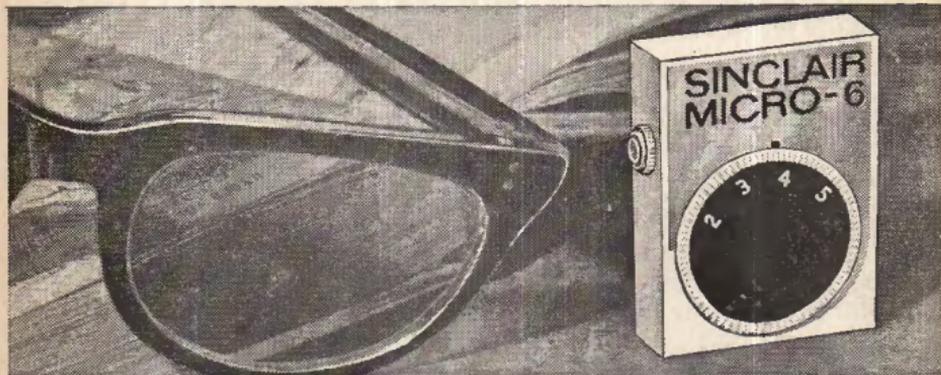
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By buying a Sinclair Z.12 Amplifier (See next 2 pages) with your Micro-6, you can save 9% and have the basis of a powerful car, portable or domestic radio. Bought separately, these items come to £7.9.0. Send now for

PACK 'A' £7.0.0

More money saving offers on next pages

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SINCLAIR RADIONICS LTD., SINCLAIR HOUSE
22 NEWMARKET ROAD, CAMBRIDGE

SINCLAIR

SINCLAIR MICRO FM

7 TRANSISTOR FM SUPERHET TUNER-RECEIVER

A 100%
BRITISH
PRODUCT



Case in black plastic with polished and brushed aluminium front

SIZE $2\frac{1}{2}'' \times 1\frac{1}{2}'' \times \frac{3}{4}''$ (74 x 48 x 19 mm.)

1

- WITH PULSE-COUNTING DETECTOR
- WITH LOW I.F. TO CUT OUT ALIGNING
- WITH 2 OUTPUTS FOR DOUBLE PURPOSE OPERATION

ALL YOU WANT FROM AN FM TUNER AND RECEIVER IN A SINGLE UNIT

For far less outlay and trouble you can enjoy building and using the most advanced F.M. design in the world when you buy the Sinclair Micro FM. This fully-fledged 7 transistor, 2 diode FM superhet has original features that ensure superb audio quality and great sensitivity together with utmost ease of assembly. There are two separate output sockets, an exclusive Sinclair feature which enables you to use your Micro FM as a tuner and as an independent pocket size portable receiver as you wish. This wonderful set is just about as big as a packet of 10 cigarettes and is as outstanding in appearance as in its performance. Many other excellent features are included in the Micro FM to make building an FM set easy for everyone. Best of all, it brings first class performance with a saving of pounds.

★ TECHNICAL DETAILS

Seven transistor, 2 diode FM superhet for use as a tuner and as a self-contained pocket receiver. The R.F. amplifier is followed by a self-oscillating mixer. Low I.F. dispenses with the need for alignment. A 3 stage I.F. amplifier produces a square wave of constant voltage which is fed into the pulse counting discriminator. This is converted into pulses to reproduce the original modulation exactly. After equalisation, the signal is fed to the audio output socket and to the receiver's own audio amplifying stage for using the Micro FM as an independent self-contained receiver. A.P.C. locks on each station tuned in. THE SINCLAIR MICRO FM is self-contained within a neat black plastic case faced by an elegantly designed front panel of brushed and polished aluminium with spun aluminium tuning dial to match.

THE WORD'S FIRST
POCKET-SIZE
COMBINED FM
TUNER —
RECEIVER



Use it as —



- Pulse counting discriminator
- Low I.F. eliminates alignment problems
- Tunes from 88 to 108 Mc/s
- Audio response—10 to 20,000 c/s ± 1 dB
- Signal to Noise ratio—30dB at 30 microvolts
- Operates from standard 9V battery, self-contained

Complete kit inc. telescopic aerial, case, aluminium front panel, dial earpiece, 2nd outlet plug and instructions.

£5.19.6

SPECIAL MONEY-SAVING OFFER

No. 2 TO P.W. READERS

To make it possible for constructors to obtain the best possible standards of reproduction, we are offering the Micro FM together with the Z.12 described opposite for £9.10.0 instead of the usual price of £10.9.0. Send now for

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SINCLAIR Z. 12

COMBINED 12 WATT AMPLIFIER AND PRE-AMP

The ultimate in
**SIZE, POWER
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 AND PRICE**

- **SIZE**—3" x 1½" x 1½" (75 x 44 x 33mm.)
- **12 WATTS R.M.S.** (24w. PEAK)
- **15-50,000c/s ± 1dB**
- **IDEAL FOR 12V. OPERATION**

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 with the Micro
 FM

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 com systems

Never has such power and quality been contained within so compact a size as that of the Sinclair Z.12. This eight transistor amplifier includes its own pre-amplifier designed to accept any type of input. The Z.12 incorporates ultra-linear class B output in a circuit using special H.F. transistors to provide exceptional performance standards. This newest Sinclair amplifier may be powered from any supply between 6 and 20 volts D.C.; the output can be fed directly into any load from 1.5 to 15 ohms. Thus the scope of the Z.12 is far greater than that of any other amplifier irrespective of price, making it the most versatile, dependable and compact audio amplifier system ever. The Z.12 is supplied ready built, tested and guaranteed, together with details of tone and volume control circuits by which input sources can be accurately matched to the pre-amp.

Complete kit of
 parts with 8 transistors, instructions and Z.12 manual

89/6

SPECIAL MONEY-SAVING OFFER NO. 3 TO P.W. READERS

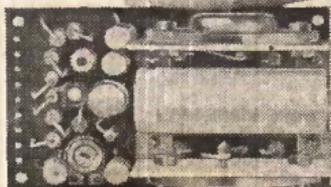
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A 100% British
 product



TECHNICAL DETAILS

8 transistor hi-fi amplifier with integrated pre-amp, using special H.F. transistors. Generous negative feedback assures superb quality.

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WEIGHT—Three ounces.

OUTPUT—12 watts R.M.S. continuous sine wave (24w. peak).

15 watts R.M.S. music power (30w. peak).

FREQ. RESPONSE—15-50,000 c/s ± 1dB.

OUTPUT IMPEDANCE—3 to 15 ohms. Two

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INPUT—2mV into 2K ohms.

POWER REQUIREMENTS—6-20V d.c.

SIGNAL TO NOISE RATIO—better than 60dB.

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If you do not wish to cut this page, please

mention PW.4 when writing your order.

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Sinclair Micro FM	£5.19.6
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1/P 230V, 50 c/s this is a Pulse Generator covering 200 c/s to 10 Kc in 2 ranges directly calibrated. The pulse width can be adjusted to 5, 4, 3, 2, 1 or 0.5 Usec, by switch on front panel. O/F pulse can be switched to +20, -10 or -10v, and is variable down to 1 Mill volt. There is also a 20 Usec, sync. O/F pulse. It is also possible to obtain a Square wave 3 to 1 and a linear sawtooth from these units, also it is possible to extend the frequency range. These are supplied in good condition with a case and outer cover, un-tested. Price £41.0.0 plus 12/6 cart.

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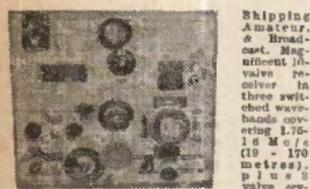
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4 VALVE 4 WATT AMPLIFIER



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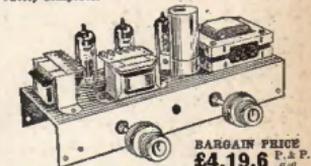
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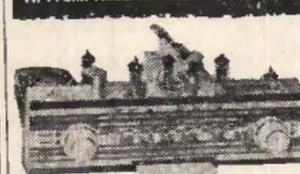
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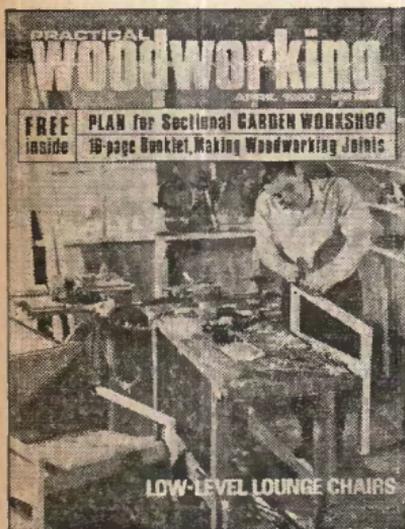
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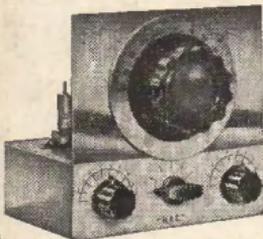
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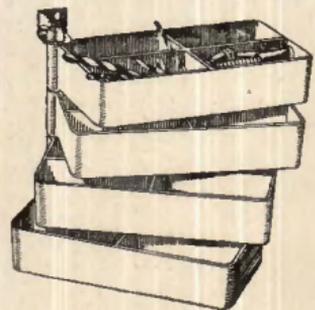
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