ess <u>`</u> and PRACTICAL TELEVISION Staff EVERY MONTH L. O. SPARK FRANK PRESTON

Vol. XIX. No. 438. DECEMBER, 1942 COMMENTS OF THE MONTH

Editor F.J. CAMM

BY THE EDITOR

1

Imported Valves—Prices Fixed

THE Board of Trade has made an Order, after con-sultation with the Central Price Regulations Committee, controlling the prices to be charged by official selling agents, wholesalers and retailers, of all radio valves imported under the Lease-Lend provisions. These prices are set out in related price lists which have been certified by the Board of Trade, the Central Price Regulation Committee and each of the Local Price Regulation Committees.

The Order provides that all agents or wholesalers dealing in these valves must supply either before or at the time of delivery to their customers a written notice which can, if desired, be incorporated in the invoice, stating the correct maximum prices for the valves in question. All retailers selling these valves must display in a prominent position in their shops a notice giving the maximum price appropriate to any of these goods offered for sale by them. This Order came into force on the September 28th, and copies are obtainable for 2d, from His Majesty's Stationery Office, York House, Kingsway, W.C.2.

The prices fixed are as follow :

Kingsway, W.C.2. The prices fixed are as follow: OZ4, 158. 3d.; 1A5G and 1A5GT, 11S.; 1A7G andIA6GT, 12S. 1od.; 1C5G, 1C5GT, 1LA4, 1LN5, 1N5G,1N5GT, 1Q5GT, 3Q5GT, 1T5GT, 5V4G, 5Y3GT/G,5Y3G, 5Y4G, 5Z4, 5Z4GT/G, 6X5, 6Z5/12Z5, 7Y4, 25Y5, $25Z5, 25Z6, 25Z6G, 25Z6GT, 35RE, <math>\frac{1}{5}$ 5Z4GT, $\frac{1}{5}$ 25Z6T, 25Z6G, 25Z6GT, 35RE, $\frac{1}{5}$ 5Z4GT, $\frac{1}{5}$ 25Z5 1C6 and 1LA6, 12S 1od.; 7B5, 7B7 and 7C5, 12J7GT, 12K7GT, 12SJ7GT, 12SK7GT, 18, 25A6G, 25A6GT, 25L6, 25L6G, 25L6GT, 35L6GT, 36, 42, 43, 47, 78, 56L6GT, 6F6, 6F6GT, 6J7GT, 6K6G, 6K6GT, 6K7, 6K7GT, 6L7G, 6SK7, 6V6GT, 12S. 1od.; 6A3. 6Q7G, 6Q7GT, 6T7G, 6SQ7, 7C6, 12Q7GT, 12SQ7GT. 75, 85, 11S. 7d.; 6A6, 6B5, 6C8G, 6F8G, 6L6G, 6L6, 6SC7, 18S. 3d.; 6A7, 6A8G, 6A8GT, 6L5G, 12F5GT, 12J5GT, 12A8CT, 12SA7GT, 14S.; 6F5G, 6F5GT, 6L5G, 12F5GT, 12J5GT, 12SE5GT and 76, 9S. 2d.; 5Z3GT/G, 6N6G, 18S. 3d.; 12A7, 25B8GT, 25A7GT, 32L7GT, 127C6C 12S 2dd.; 22A, 124, 125GT, 127C6C 12S 2dd.; 22A, 21, 125GT, 12SE5GT and 76, 9S. 2d.; 5Z3GT/G, 6N6G, 18S. 3d.; 12A7, 25B8GT, 25A7GT, 32L7GT, 12SC7GT, 125A7GT, 32L7GT, 12SC7GT, 125A7GT, 32L7GT, 12SC7GT, 125A7GT, 32L7GT, 12SE5GT and 76, 92. 2d.; 12SG7GT, 125A7GT, 32L7GT, 12SE5GT and 76, 92. 2d.; 12SE5GT and 76, 93. 2d.; 12SE5GT and 76, 93. 2d.; 12SE5GT and 76, 93. 2d

25B8CT, 25A7CT, 32L7CT, 25B8CT, 25A7CT, 32L7CT, 117Z6G, 15s. 3d.; 12Z3, 11s.; 45, 11s. 7d.; TZ40, 66s. 11d.; 6SF7, 15s. 3d.; 6J5CT, 9s. 2d.; 6J8C, 11s. 12B8CT, 70L2CT 618G, 14s.; 12B8GT, 70L7GT and 83, 15s. 3d. This new Order regularises the position, and will prevent the racketeering in American valves in which a few unscrupulous dealers have engaged.

The Position of Service Engineers

THE Radio Manufacturers' Association has circularised all dealers (who filled up their ques-tionnaire on the subject), trade service organisations, manufac-turers and wholesalers, informing them of the decision of the Ministry of Labour on the position of radio service engineers, in relation to deferment. The R.M.A. by means of its questionnaire has collected data relating to the needs of the has collected data relating to the needs of the public in connection with the servicing of receivers. About 4,000 dealers filled up the questionnaire. As a result negotiations have been in progress between the Ministry and the R.M.A. on the question of re-servation of a sufficient number of men to maintain an efficient public service. The importance of radio during the war has been stressed, for more than ever it represents the official voice and the official mouthpiece. Until the negotiations had been completed the Ministry had left the industry alone, and had not called up service engineers. The R.M.A. letter says that apart from the question of principle the scheduling of service depots under the Essential Works Orders would present difficulty. Inder the Essential works orders would present dimetary. It has not been possible to agree on any general scheduling of the wireless service industry. If any dealer has a separate repair depot of any size, particularly if it is doing military work, he can apply for scheduling either through the Supply Department concerned, or if he is on purely civilian work, through the Ministry of Information. Applications for the deferment of service men who are essential, and cannot be replaced by older men, or by wemen, should be made in the usual way. The applica-tions should state the average number of repairs, both major and minor, carried out per month during the past r2 months, the number of persons the department is divided into (skilled and semi-skilled) and if the area served is limited, any general conditions, relating to the adequacy of repair facilities in the district.

The Ministry of Labour is not willing to work on any rigid datum line as to the number of repairs per person, i view of varying local condi-tions and uncertain demands on

man-power. The Committee con-cerned with these negotiations suggested that retailers should see subsection that not retained should see that not-less than 55 repairs per skilled man with 25 per cent, addi-tion for each semi-skilled worker are carried out each month.

If deferment is refused the assistance of the Regional Representa-tives of any Government Department interested in the work of the firm, generally the Board of Trade, should be obtained if it is considered that an exceptional case could be made for the deferment of the men concerned. Such assistance should not be sought merely because the Man-power' Board has rejected the application, for the Board will have already consulted the appropriate department before deciding to reject the application. Women born between December 7th, 1910, and December 31st, 1922, will not be withdrawn at present without prior substitution;

We are of the opinion that the radio industry has made a generous contribution of man-power to the Services, and should be left alone

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our research is south with the latest develop-ments for give no verranity that apparatus-described in our columns is not the subject of lifes published in Practices, by the subject Expeription of the research of the subject articles published in Practices. WINNERS is signatury to the forme Concention and the U.S.A. Repredictions or initiations of any of these are therefore expressly forbidden. Fractices, WINNERS in a conditions are diverticed in this paper should not be taken as an indication that they are necessarily available exercise.

as an indication in available for export,

December, 1942



Sirens on Radio Sets

2

A NEW YORK firm has invented a private siren, to be attached to a household radio set, which will give the alarm six or seven minutes before the official warning. The set, which is left switched on, starts to howl when the radio stations go off the air.

"Calling All Cars" **I** T is reported that the American system of calling all cars by radio telephone, instead of one-way Morse, is to be adopted by Socilard Varie abortive. Tarabaical **L** cars by radio telephone, instead of one-way Morse, is to be adopted by Scotland Yard shortly. Technical experts at Scotland Yard have perfected a two-way wireless system which will enable the information room staff to converse with the crews of all Metropolitan police cars within a radius of 75 miles. There is no possibility of anyone else listening in.



The first radio sound effects woman in America, Miss Beth Msir, in action. She produces the sound effects for the General Electric Company's frequency modulation playhouse.

Radio Here for U.S. Troops

THE broadcasting of special programmes from the United States direct to the American forces in Britain has been the subject of discussions between the Britan has been the subject of discussions between the B.B.C., the United States Office of War Information, and American Army and Navy chiefs in London recently. It is likely that "home-town news" for U.S. troops will be an important feature.

Radio Salary Cuts!

A CCORDING to a recent report, Dr. James Thomson, new general manager of the Canadian Broadcasting Corporation, is to be paid £10,980, including expenses, in place of £12,000, and Major Gladstone-Murray, the Director-General, will receive £12,000 in place of £14,000.

New Bulgarian Stations

T is reported that seven new transmitting stations are being built in Bulgaria, presumably under German instructions, one of which will be the most powerful in the Balkans.

Heard Herself Broadcast

A^N Englishwoman recently had the novel experience A of recording in England a radio message to her husband, a warrant officer in the R.A.F. in Canada, and husband, a warrant oncer in the K.A.F. in Canada, and then hearing it broadcast while sitting beside her husband in a house in Saskatchewan. Between the recording and broadcasting she had crossed the Atlantic by ship. Shortly before she sailed she was invited to take part in a B.B.C. greetings feature.

Unlicensed Radio Sets

DURING a recent prosecution at Liverpool, on behalf **D** of the Postmaster-General, when a number of persons were fined f3 each for not having radio licences, it was stated that in cases where the Postmaster-General considers it necessary he will exercise his right to seek

confiscation of the apparatus.

U.S. Radio Curfew

I^T is reported that the majority of American radio stations are to close down at midnight in order to save fuel.

Receiver Sales in Canada

THE Dominion Bureau of Statistics states that the sales of radio receivers in Canada during the first three months of this year amounted to \$7,953: During the previous quarter the figure was 118,647, and 61,038 in the corresponding period of 1941. The greatest increase was recorded in the province of Ontario.

"Uncle Arthur"

I^T is interesting to note that Mr. Arthur Burrows, the first B.B.C. wireless announcer, has been appointed Director of the Broadcasting Division of the Ministry of Information: Few men have had a wider or more varied experience of broadcasting than Mr. Burrows, whom radio enthusiasts will remember as "Uncle Arthur" in the 2LO days. With Mr. Stanton Jefferies, he was in charge of the first B.B.C. programme broadcast on November 15th, 1922. For several years before the war he was manager of the International Bureau of Broadcasting at Geneva.

New B.B.C. Appointment for Fred Hartley

FRED HARTLEY has been appointed Light Music Supervisor of the B.B.C. He was born in Dundee in 1905, began to learn the fiddle when he was four, and In 1995, began to learn the hiddle when he was four, and the plano at six, because his father did not want him to continue with the fiddle. He won various awards and medals, played at hundreds of concerts, came to the Royal Academy of Music at 16, won the Sterndale-Bennett Scholarship and was official accompanist of the academy academy.

He says he "wasted a lot of time," and spent many hours plaving in dance bands—the London Collegians, Jack Hylton's Kit Kat Band, and was on and off the music hall stage for two years.

A stormy interview with his academy principal ended his studies, and he became conductor at the biggest music-hall in Stockholm, after which, in 1926, he returned to this country to become accompanist to the B.B.C. in the Dundee studio.

Five years later he came back to London to form what

was originally known as " Fred Hartley and his Novelty Quintet." He was the first person to use singers as an integral part of a light music ensemble. It was in this way that he introduced to radio Cavan O'Connor, Brian Lawrence, Webster Booth and others.

In his own programmes he does almost everything himself, even to writing the scripts; for a long time he also ran a music publishing firm, and he still keeps an eye on a hand-laundry business in the country. His versatility is equal to his energy.

Light music, he believes, is as essential as variety, and even moderately good light music is better than bad variety. He hopes to improve the standard of light music. Moreover, the kind of light music that suits the early morning does not suit the late evening, and he is keen on the better adaptation of a programme to its period.

B.I.R.E. Meeting

A^T a members' meeting of the British Institution of Radio Engineers, held on October 23rd, a paper on "Theory of Units" was given by L. H. Bedford, M.A., a synopsis of which is given below :

In the formulation of electro-magnetic theory there are three stages at which arbitrary constants are conveniently introduced for the purpose of defining units. It is customary to assign immediately the value unity to certain of these constants, after which they are lost sight of, and have often, at a later stage, to be painfully resuscitated. A sketch of electro-magnetic theory was given in which the suppression of these fundamental constants (k_1, k_2, k_3) did not occur. Maxwell's theory was shown to lead to a certain relationship between the k's involving the velocity of light. Subject to this restriction, the assignment of k-values is an arbitrary matter, the process of formulating a unit system being one of two degrees of freedom.

A table was constructed showing the assignment of k-values corresponding to the various known unit systems. This table can, among other things, be used to derive the numerical relationship between the unit quantities of the various systems. One necessary example of this was given, but it was pointed out that the (advocated) use of a generalised system of units (k's unspecified) eliminates once and for all the tiresome and inclesant process of unit changing. Moreover: this and inelegant process of unit changing. Morcover, this generalised system of units is free of the dimensional inconsistencies which characterise the known systems. at least in their more usual modes of expression.

An example was given of the direct use of the generalised system of units; a particular problem which would normally present the most tiresome process of unit changing was solved straight out in generalised units from which the numerical answers were written down at once in practical units. Serious inconsistencies of method and nomenclature connection with the in practical system of units were brought to light, and proposals for rationalisation considered.

Broadcasts to Schools RADIO'S growing impor-tance in the educational world is shown by the in-creasing number of schools registering for the special B.B.C. broadcasts. It is hoped that before the end of the school year, which recently opened, the number of listening schools will have reached the

12,000 mark. Already 10,429 schools have registered, which is 469 higher than at the beginning of the last The 1941-42 year ended with a total of school year. 11,299, a figure nearly 2,500 in excess of the highest figure recorded at a corresponding date in any previous year.

It is significant that the figures have risen in spite of the upheaval caused by the war. In pre-war years, for instance, the total was 6,847 at the end of 1936-37, and for 1938-39 registrations numbered 9,952. Incidentally, it should be noted that the above figures do not include Scotland, where a new system of registration has been introduced, and the complete returns are not yet available.

Factors contributing to the success of last year's broadcasts included the issue of an attractive annual programme, the co-operation of local education authorities, and, finally, the extent to which broadcasting has become an accepted part of school life.

New U.S. High-power Station

THE General Electric Company of New York recently opened a roo-kilowatt radio station (WGEO) to be used for broadcasting news to "the oppressed people of the world." In the opening broadcast the announcer said to the people of Germany: "We are coming with a cloud of 'planes that will darken your land, and blast and destroy your cities.'

Radio Telephone to E. Africa

A RADIO-TELEPHONE service between Portugal and Mozambique, Portuguese East Africa, is to be opened shortly.

Incorporated Radio Society of Great Britain

WE have been notified by the General Secretary that the following persons have been nominated to serve on the 1943 Council of the above society :

Officers: President, Mr. A. D. Cay (G6NF); Execu-tive Vice-president, Mr. E. L. Gardiner (G6GR); Hon. Treasurer, Mr. A. J. H. Watson (G2YD); Hon. Secre-tary, Mr. H. A. M. Clark (G6OT); Hon. Editor, Mr. J. W. Mathews (G6LL).

Mathews (GoLL). Members (retiring members eligible for re-election): W./Cmdr. G. M. R. Scott Farnie (GW5FI), Mr. G. A. Jessup (G4HG), Mr. W. A. Scarr (G2WS), Mr. E. H. Simmonds (G8QH). New nominations as per Article 43: Mr. F. Charman (G6CJ), Mr. D. N.*Corfield (G5CD), S./Ldr. John Hunter (G2ZQ).



Empire pilots taking a course of Morse signalling at an R.A.F. Army Co-operation Command.

Short A Superl let

An Interesting Layout for the Experimenter. By R. L. GRAPER



HE following description is of a superhet layout built by the writer primarily for experimental work on the short-wave band. The circuit, Fig. **r**, shows a five-valve receiver incorporating a hexode frequency-changing valve followed by an I.F. amplifier, second detector, resistance amplifier, and pentode output. Interesting features are as follows : The coils are wound Interesting leatures are as follows: The coils are wound on rin, formers, and plug into sockets, which are mounted on stand-off pillars, and are easily changed for various wavelands, Figs. 2 and 3. The oscillator coils have small trimmer condensers soldered to the pins of the grid section, Fig. 4. The I.F. transformers are mounted on stand-off feet, to simplify wiring. An interesting point regarding these I.F. transformers

Thick

.....

Figs. 2, 3 and 4.—Construc-tional details of coils and holder.

is that they were originally made for a frequency of 110 kilocycles, but were converted for 465 kilocycles working by careful removal of an equal portion of each winding.

I.F. Transformer Conversion

The conversion of the 110 kc's L.F. transformer to 465 kc/s was carried out as follows : The spool carrying the two coils was removed, after first observing and marking all leads. The spool was then mounted on a spindle in a vice, and a slight braking action obtained by rubber washers. Both coils had turns removed (Fig. until equal to the gauge card, which was marked with the thickness of coils, and cut-out measured for this. Approximately three-quarters of their thickness was removed. would suggest removing two-thirds, and testing by means of a known frequency liner, which is facilitated by the method of mounting the trans-

(Left) Above chassis view of the 5-value superhet, showing the layout of components :

1. Aerial stand-off	7. L.F. value.
insulator.	8. Earth wire,
2. Aerial coil.	9. Pen. 220.A.
3. TH 2.	10. I.F. fect.
4. VP 2.	11. Volume control
5. Oscillator coil.	and switch.
6. Trimmer condenser	12. Epicyclic Drive
soldered to grid	13. PM 2 H.L.
pins on each coil.	

Bakelite 22 Insulating Pillars



4

December, 1942

Plywood covered

with metallised paper.

former on stand-off feet. As trimmers were found to need adjusting to almost three-quarters maximum, less turns might be removed with advantage. No attempt was made to shorten the distance between coils, to produce closer coupling, but this might be found advantageous, if found possible, in other designs of 110 kc/s transformer.

Aerial and Oscillator Condensers

The aerial and oscillator tuning condensers are not ganged, although by rearrangement of layout they could be converted to single control, as the two condensers were found to keep well in step over the waveband covered. By the way, the oscillator condenser is about half the capacity of the aerial tuning condenser, being double spaced as regards the vanes. Both acrial and oscillator condensers are approximately of .ooor mfd. maximum capacity, and .00005 mfd. respectively. Note the main earth wire, raised just clear of the metal chassis, for direct earth connec-



This is also tions. incorporated, on the underside of the chassis. The valve holders for the hexode frequencychanging valve, and the one for the I.F. amplifying valve, are raised clear of the chassis for short direct connections to the coils and condensers.

Coil Data

There is one point which should be noted with regard to the construction of the coils and holder, as it will save a certain amount of time and labour. Although two centre pins are shown, it is sufficient to provide and use only one, as their connections are actually common, i.e., both taken to carth. The ultimate position of the pin is determined according to the number of turns.



Fig. 1.—Theoretical circuit diagram.

12" 8 Centres vary owing to different brackets utilised \$ Sq

Fig. 6 .- Panel layout.

Coil No. 1 covers an approximate wave range of 14 to 25 metres. Its oscillator anode winding consists of 12 turns of 20 s.w.g. and the grid 10 turns of 18 s.w.g. The aerial coil primary requires 8 turns of 20 s.w.g. and the grid coil 10 turns of 18 s.w.g.

Coil No. 2 has an approximate werage of 30 to 57 metres. The oscilcoverage of 30 to 57 metres. The oscil-lator anode section requires 20 turns of 32 s.w.g., and the grid coil 27 turns of 22 s.w.g. The primary of the aerial coil has 16 turns of 32 s.w.g., while the grid section consists of 27 turns of 22 s.w.g.

General Construction

The photograph of the completed

step by step, that individual parts of the circuit are approaching 100 per cent., than to build up an elaborate design, and, if one is unlucky, have to cut out sections here and there to try and find the cause

plan. Dimensions for the panel, as used by the writer, are given in Fig. 6.

The construction lends itself to experimental work, which, as mentioned earlier, was one of the main objects -the writer had in mind. Unfortunately, it was not

possible to put the receiver through exhaustive tests—owing to service in the Forces but sufficient results were obtained to prove that the general design that the general design and layout were capable of putting up a very satisfactory performance which, no doubt, could even be improved upon as soon as time permits further tests and adjustments.

It will be noticed that no attempt' has been made to incorporate some of the refinements which are usually embodied in a super-super or a "communica-tion" receiver. Such items can be added at a later date, according to the components and valves available. During the initial experimental work it is advisable to keep the circuit as straightforward as poscircuit as sible, to eliminate likely sources of trouble. It is far better to improve and extend, making sure,

	COMPONENT
	VALUES
Cl	.1 mfd.
C2	.0001 mfd.
C3	.0001 mfd.
C4	.1 mfd.
C5	.25 mfd.
C6	.1 mfd.
C7	.0001 mfd.
C8	.25 mfd.
Ċ9	.01 mfd.
C 10	.01 mfd.
C11	.002 mfd.
R1	5,000 ohms.
R2	50,000 ohms.
R3	30,000 ohms.
R4	50,000 ohms.
R5	1 megohm.
R6	30,000 ohms.
R 7	20,000 ohms.
R8	1 megohm.
R9	100,000 ohms.
R 10	0.5 megohm.

WIRING DIAGRAM OF THE 5-VALVE SUPERHET

of some particular trouble.



Fundamental Oscilloscopic Tests

Voltage and Current Measurements, Phase and Frequency Comparison

3:2

With linear time base

4:1

Fig. 2.- Typical figures for frequency ratios. These are known as Lissajous figures.

By S. A. KNIGHT

T is not possible in an article of this nature to go over the use of the oscilloscope in every field of its applicaof basic tests, and uses which are invariably required in almost every-application of the instrument. Generally, tests which are carried out by the oscilloscope can be divided into voltage and current

to be obtained.

on these fundamentals considered

2:1



readings.

Voltage Tests

Since the application of a voltage to either pair of deflecting plates will cause the spot to move across the screen, it is a simple matter to see that by measuring the amount of movement caused. the potential of the applied voltage can be determined. In this way the cathoderay tube becomes a voltmeter which will read for either A.C. or D.C.

The difference, however, between an A.C. and a D.C. measurement is that in the case of the former the spot is drawn out into a line the length of which is equal to the peak voltage applied, while in the latter the spot just changes position, and finishes up in a different position on the face of the screen.

It is, of course, necessary that before a tube can be used in this way its sensitivity must be known, and the screen accordingly calibrated. Most tubes are given certain figures for sensitivity by the makers, and this is usually quoted in so many volts per centimetre deflection. A celluloid graticule divided into cm, squares, and also preferably millimetre squares, attached over the face of the screen is therefore necessary.

For obvious reasons, an oscilloscope is rarely used for D.C. measurements, but for A.C. work it is extremely useful. It is practically independent of the frequency of the supply under measurement, the input impedance is generally quite high and loading errors are avoided, and the scale is linear, i.e., a trace will double its length if the applied voltage is doubled.

Since the oscilloscope measures peak voltage, this must be converted into R.M.S. by multiplying by .707 or dividing by 1.414. No time base is required for voltage measurements, a single vertical trace being sufficient for measuring purposes.

Current Measurement

Current readings can be taken similarly to that of voltage by using deflector coils mounted on the outside of the tube. For vertical deflection two coils placed

one on either side of the tube neck are required, such the use of the oscilloscope in every field of its applicathat a horizontal field is produced when a current is tion, but a certain amount can be said on the subject flowing through them. (Fig. r.) The electron stream travelling along the fube axis will be deflected up or down while it is under the influence of this magnetic field, the magnitude of the deflection depending upon the current in the coils.

The screen can therefore be calibrated for current measurements, larger currents than those producing full screen diameter being read by the method of conventional shunting.

It is sometimes more convenient in practice to measure current electrostatically, that is, by means of the deflector plates and not coils, since inductive effects of the latter are sometimes such that the constants of the circuit under examination are upset and errors introduced.

The voltage dropped across a non-inductive resistor

in series with the load circuit will give a measure of the current flowing, provided such drop is small compared with the total supply to the test circuit. This method generally requires the use of amplifiers. to produce a satisfactory deflection, and this has its disadvantages.

In all cases where amplification is necessary, it must be borne in mind that frequency range is probably limited to certain maximum and minimum figures, and that most amplifiers impose a load on the system under measurement.

Lissajous Figures and Frequency Comparison When the ratio of the

time base frequency and an applied work frequency can be expressed as an integer,

With two

sinusoidal

inputs

te, a whole number or a ratio of two whole numbers, the resultant trace is a stable figure. This might be a single or an intricate pattern which might at first sight appear difficult to understand, but which is quite simple of interpretation.

3:2

If the ratio of the two frequencies is not an integer, but has considerable differences, the pattern races and is confused. Typical figures obtained for various ratios are shown in Fig. 2, these being known as Lissajous figures, and the determination of the ratios is arrived at in the following way.



The ratio of the number of complete peaks in the horizontal direction to the number of entrants in the vertical direction, or the ratio of the number of complete peaks in the horizontal direction to the number of horizontal lines joining similar points of intersection of the pattern waves, plus one.

A simple method of frequency comparison up to ratios of about 12: r is obtainable in this way, using a known frequency applied to one pair of plates, and the unknown applied to the other pair (Fig. 3). As variation is made in the unknown frequency, Lissajous figures will appear as stable patterns at certain settings, and the ratio can then be easily determined between the two in Fig. 4 was obtained, then if the 'known frequency was taken from 50 \circ mains, the unknown is obviously 200 cycles, since their ratio is 4: r.

Careful counting of the loops is necessary where the ratio becomes high, and generally above ratios of 12:1 the pattern becomes too complicated for the Lissajous rule to be applied. Other methods must be used, which will be described later.

Phase Shift Determination

The cathode-ray oscilloscope is useful in checking the phase relationship between two alternating voltages. The phase relationship at various points in an amplifier, or the phase distortion in a complete amplifier, can be determined, the difference in phase being shown by the position of the trace on the screen.

A simple circuit is shown in Fig. 5, where the phase

distortion in a complete amplifier is being examined. By suitable adjustment of R_1 and R_2 , relative amplitudes of X and Y deflection may be equaised, and the figules on the screen easiry inspected.

Als will be seen from Figs. 6 and 7, as the phase difference moves from o to 180 deg. a line trace is formed moving from an angle of



45 deg. to the left, through an angle of 90 deg. back to a line of 45 deg. to the right, and changing from a line to an ellipse and a circle as the phase passes through the intermediate stages. Then at any position the phase angle is equal to the angle subtended by the axis AB at



the vertex C, i.e. : Tan $\theta/2 = AB/CD$ (Fig. 8 a) and assumes that the X and Y amplitudes of deflection



Fig. 7.—The line traces formed as the phase difference moves from 0 to 180 deg.



Fig. 8.—(a) Tan $\theta/2=AB/CD$. (b) Sin $\theta=AB/AC$.



Alternatively (Fig. 8 b) the phase angle 9 are equalised. is given by:

 $\sin \theta = AB/AC$

Phase Angle of a Reactive Load

This can generally be determined by using a circuit as in Fig. 9. . If by suitable adjustment of resistance R the relative amplitudes of the X and Y deflections are equalised, then



Impedance of load =Z=R $= R_{L}^{2} + X_{L}^{2}L$ Axes ratio = $\tan \theta/2$

where $\theta = \text{phase angle}$ of the load.

Determination âf Capacity by Phase Pattern

The circuit shown in Fig. 10 will, if the reactance of the condenser C is equal to resistance R, give a circular traverse, as there is a 90 deg.

phase shift across a capacitive load. If an unknown capacity is connected at C and R is adjusted so that a circular trace results, then :

$$R = \frac{r}{WC}$$

and C can be determined knowing the resistance of R.

Phase Delay

Phase shift, measured by the ellipse method, can be converted into phase delay, the expression giving the time in seconds required by a cycle of a certain given frequency to travel through the wiring of a receiver or amplifier under test.

ase delay=
$$\frac{\text{phase angle}}{\text{Frequency} \times 360}$$

Phase distortion occurs when the delay in time varies with the frequency injected. If there is no variation for all the frequencies handled then there is no phase distortion, and the phase shift is proportional to frequency. This method of investigating phase distortion is used chiefly in television work, where trouble of this nature can prove objectionable.

Precautions Necessary

In studying phase shift patterns on the oscilloscope, it must be remembered that the X and Y deflector plates generally have different sensitivities. The inclination generally have different sensitivities. The inclination of the line (0-180 deg.) will not therefore be at 45 deg. to the horizontal even if the inputs are equalised, neither will a 90 deg. shift in one produce an exact circle.

Again, apart from the above, if the input voltages are



Fig. 9.—Circuit to determine phase angle of a reactive load.

it must be seen that this itself does not introduce a phase shift. Single stages of amplification hearly always cause a shift of 180 deg.—apart from cathode followers which are not actually amplifiers—and this must be watched and allowed for.



Other Methods of Frequency Comparison

When the ratio between the work and sweep frequency exceeds r2:r, a limit is reached such that comparison by simple Lissajous figures is impossible. Other arrangements are therefore made.

One method is to apply the lower of the two frequencies in and out of phase circuit similar to Fig. 10 to the X and Y axes, thus producing a circular trace, and then modulate the tube grid by means of the other frequency, such that the beam is extinguished at peak amplitudes and the trace appears as a dotted ring.

The ratio of the frequencies is then given by the number of dots. If the frequencies differ considerably the pattern races and is confused, but slip frequencies ot a few cycles will show as a slow rotation. (Fig. 11.)

It might happen that adjacent dots do not represent successive cycles, i.e., the frequency ratio is not an integer, and dots on the second rotation of the circular frequency may be occurring between those of the first one. If a magnet is moved quickly across the tube, such that the beam is moved horizontally, the dots will move as a line if the frequency ratio is an integer, and will break up into separate rows if it is not so.

Another method is to apply one of the frequencies as a circular trace as above, and apply the other to the Y axis. A figure similar to Fig. 12 will result where the ratio can be determined by counting the number of peaks occurring on the circle.

The Meaning of Waveforms

The most useful feature of the oscilloscope is its ability to produce actual waveforms of voltage and currents occurring at various parts of a circuit. A linear time base must be applied to the horizontal controlling plates, while the waveform to be observed is applied to the vertical controlling plates, either direct or through an amplifier.

The ratio between the time base frequency and the

not equal in amplitude, the inclination will be different, and the tangent of the angle of inclination to the horizontal will give the ratio of the vertical to the horizontal voltage.

Two voltages, though out of phase, may be compared in this way.

If it becomes necessary to use an amplifier in the oscilloscope, waveform to be observed gives the number of separate waveform images observed on the screen; thus to give the image of a single wave the frequency of the time base and the signal must be identical.

For obvious reasons it is not possible to give details of every type of waveform which can be observed on an oscilloscope, the number being practically limitless, and the type depending on conditions. A few examples are given in Fig. 13, however, and show how actual circuit conditions can be represented on the screen.

- Trace A. Undistorted sine wave, applied, say, to the grid of a valve.
- Trace B. Output of the valve, showing distortion due to grid exceeding cut-off.
- Trace C. Squegging action, due to grid current charging the grid condenser of a valve and cutting the valve off. The exponential discharge through the grid leak is shown. (Grid waveform.)
- Trace D. Anode waveform of the above valve, showing the negative pulses which occur during the period when the valve is conducting heavily.
- Trace E. Continuous oscillation in a circuit.
- Trace F. Intermittent oscillation.



Trace G. Half-wave rectifier output without smoothing.

Trace H. The above with conventional smoothing arrangements.



Fig 13.-A few examples of the many types of waveform produced on the oscilloscope.

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Ground Direction-finding

How the Bearings and Positions of Ships, Aircraft and Other Transmitting Stations can be Determined from a Fixed Receiving Station

MANY readers will remember the article on directionfinding from aircraft, which appeared in the issue of PRACTICAL WIRELESS dated December, 1941. It was explained how a rotating loop aerial picks up strongest signals from a transmitter in line with the plane of the loop, and how the pick-up is zero from a station on a line at right angles to that plane.

A similar system of D.F. could be used on a ground station, but there are many objections to it. One of the greatest of these is that a very large loop aerial would be required to



the low-power transmitter used in the average aircraft, for example. Another is that a rotating loop aerial mounted near to the ground is subject to appreciable errors. In any case, the mechanical operation of a large loop would present a number of difficulties. Bellini-Tosi System

strong signals from

sufficiently

receive

In practice, a very simple system is employed, and one which is capable of producing bearings accurate to a very high degree. One of

Fig. 1.— The connections for a rotating loop aerial, as used for direction-finding in an aircraft.

the first systems to be adopted—and it has been in use for many years—is that known as Bellini-Tosi. The general arrangement is shown in Fig. 2, and this should be compared with the method employed on aircraft, illustrated in Fig. 1. There are two large square or rectangular loops, fixed in position and mounted at right angles to each other. Normally, these are mounted so that the plane of one is in a north-south line, whilst the plane of the other is from east to west.



Fig. 3.—How a goniometer is made, and how it is connected to the input circuit of a receiver.

The Goniometer

A fixed inductance coil is connected in the lower arm of each of the fixed loops, the two coils being mounted at right angles to each other and wound on a cubical former, as illustrated diagrammatically in Fig. 3. Within these two fixed coils is a third coil, this one being so mounted that it can be rotated within the fixed coils. It is electro-statically screened from the fixed coils. It is electro-statically screened from the fixed coils but arranged so that it can be coupled inductively to either or both of them. Clearly, when it is parallel to one of the coils it is closely coupled to it, so that any H.F. circulating in the circuit of the fixed coil will be induced into it. Its coupling with the other fixed coil



Fig. 2.— The essentials of the Bellini-Tosi system of directionfinding. The fixed loops act as D.F. aerials, and these may be several yards square.

is nil. The complete component, comprising the two fixed coils and the rotating coil, is known as a goniometer or radio-goniometer.

The pivoted coil is described as the search coil, and is connected to the input circuit of a receiver, through a separate winding on the receiver tuning coil. Suppose, now, that the receiver is tuned to a transmission which emanates from a point due north of the aerial system. The transmitted signal will be received at maximum strength when the search coil is rotated so that it is parallel to the coil in series with the north-south loop. The signal will gradually be reduced in strength as the search coil is turned through 90 degs, and when the search coil is parallel to the coil in the east-west loop the signal will be inaudible. In practice, the signal may not be eliminated altogether, but it will, at any rate, be of reduced strength.

Should the rotation of the search coil be continued, the signal will again increase in strength, once more reaching a maximum when the search coil is again parallel to the north-south fixed coil. Similar results would be obtained if the themselves

ly connected.

Array

Action of the Goniometer

As with a rotating loop, it is necessary to "search" for the position of minimum signal strength, because this is more defined and more easily detected by the ear than is the position of maximum signal. It is possible to show that if the receiver is tuned to a transmission emanating from the



Fig.4 .- A more practically convenient type of loop aerial assembly than that shown in Fig. 2. The whole system can be sup-ported by means of a high central mast and fourshort masts, one at each corner.

present rectangular loops various large fixed rectangular loops present van mechanical difficulties in regard to crection. In nection therefore, the arrangement is often sim-plified, by making the loops triangular in shape, as shown in Fig. 4. They can then be erected by using a central-high mast, with four shorter masts, one at each corner of the aerial array. There are no serious objections to this buyers and it has been in the shorter of the second stress of the to this layout, and it has been widely employed.

Adcock Aerials

In the ordinary way, both layouts described are liable to certain errors, especially on high frequencies. These are due to the horizontal arms picking up horizontally -polarised waves, which may result from reflection and refraction of the radiated signal. This difficulty is perhaps most readily overcome by the use of a so² called Adcock (due to its originator) aerial system, types of which are illustrated diagrammatically in Fig. 5. The essentials of this array are that there are four vertical aerials situated at the corners of a square. The leads from these are generally either "overlapped," or screened, to avoid pick-up by the horizontal leads.

One practical form of Adcock layout is shown in Here it will be seen that there are four Fig. 6. self-supporting or mast-type vertical aerials placed at the four corners of a square whose diagonals are in north-south and east-west directions. These aerials are mounted on insulating supports, and the leads from them are by co-axial or similar low-loss screened cable. The cable is taken underground to the receiver cabin, which is situated in the centre of the aerial system, and the screening is adequately earthed by being connected to substantial buried plates.



Fig. 5.- These types of Adcock fixed D.F. aerial, where one winding right-angles to that shown.

The "Sense" Aerial

The four low-loss leads are connected to the fixed The four low-loss leads are connected to the fixed coils of the goniometer, as previously explained. In addition, there is a fifth aerial, this being the "sense" aerial. For convenience, and to keep the array com-pletely symmetrical, it is suspended from four rope supports attached to the upper ends of the vertical D.F. aerials. In planning this—and all other types of D.F. aerial system, the utinost care is taken to ensure complete symmetry; any out-of-balance will render the bearings taken inaccurate since the pick-up and/or losses in the four arms will vary.





Transmitter

Figure of 8

12

Aerai Foiar Diagrams It has already been pointed out that rotation of the search coil will give two positions of minimum signal, and that it is necessary to find the direction to the transmitter. This is known as "sensing," and is dependent upon the same laws as those governing the use of a rotating loop aerial. If we were to draw the polar diagram representing the signal pick-up by one pair of DER aerials we should obtain a figure of eight as shown D.F. aerials, we should obtain a figure of eight, as shown by broken lines in Fig. 7. This explains why there are two positions of the search coil for minimum and maxinum signal; the minima are obtained when the coil is set north and south, and the maxima when it is in an east-west position, as represented in Fig. 7.

But by connecting another aerial (the so-called vertical or "sense" aerial) this diagram can be modified. The The polar diagram of this aerial is a circle-since the aerial has equal pick-up from all directions-as represented by the chain line in Fig. 7. The effect of combining the pick-up from both sets of aerials is to produce a heart-shaped or cardioid diagram such as that indicated by a full line. It will be seen, therefore, that when both aerial arrays are in use there is only one direction from which a minimum signal can be received.

We can make use of this property by providing two pointers for the search coil, as also illustrated in Fig. 7. They are 90 degrees apart, and one of them is for D.F. and the other for "sense-" Assuming the transmitter to be in the direction shown in Fig. 7—due west of the aerial array—the procedure for taking a bearing would be as follows. The search-coil knob would be rotated until a minimum-signal position was found ; this would be at either o degrees or 180 degrees, as indicated by the D.F. pointer. Suppose the pointer were at zero degrees.

"Sense " Determination

We should then bring the "sense" aerial into use by closing the on-off switch shown in Fig. 3. Next, we should swing the search coil through 180 degrees and back again. If signal strength were at a minimum when the coll was in its original position, the direction of the transmitter would be as shown by the "sense" pointer. On the other hand, if maximum signal strength were obtained at that setting the pointer would show the reciprocal of the bearing.

The procedure briefly explained is not necessarily that which would be followed in all cases, since the precise method of operation would depend on the actual design of the goniometer and the arrangement of the pointers. But the principle does not vary. Should any reader



Sense Aerial Pick-up

Fig. 7.—These diagrams show how "sensing" is carried out after finding a minimum signal with the D.F. aerials.

Cardiord

350 340

experience any difficulty in following the general tech-nique, it is suggested that he might look up the article on aircraft direction-finding which has previously been referred to.

So far we have considered only the method of finding the bearing of a transmitter. Should it be necessary to establish its position it would be necessary to employ two or more D.F. installations and to take a bearing from each at the same time. If those bearings were then plotted on a map, the point at which the two or more bearing lines intersected would give the exact position of the transmitter.

W.A.A.F. Wireless Operators

AN aircraft is flying somewhere over Germany. A N aircraft is flying somewhere over Germany. It has flown through the flak, delivered its load of bombs and got back once more through the barrage. Now the pilot wants further instructions or assistance, and it is the job of his wireless operator to get them. The transmitting set has been damaged by a fragment of German shell, and the operator is not quite sure that his signals will be strong enough to be picked up at all. However, he perseveres, tapping out his message at intervals, and hoping it will get through.

Back at the station sits an airwoman, wearing earphones, listening intently for any signal that may come on her wavelength. She moves the "knob" in-finitesimally one way and another, to make sure that she is tuned in correctly. She has had nothing through for an hour and a half, but she cannot relax for a second, because at any moment a message might come. Now, as she strains here ears, she can hear the faintest of sounds. Surely they are dots and dashes. Yet they are so faint that she cannot make sense of them. She listens still more intently, until she detects some sort of rhythm in the signal. At last she is able to piece

together a connected message. She writes it down and passes it on.

A decision is taken and another airwoman begins to transmit signals with a firm rhythmic touch. sage flashes to its correct destination, passing perhaps through various links to a remote part of the country. On the accuracy of these W.A.A.F. wireless operators

in sending or picking up messages the success of a whole operation may depend. They work in watches, whole operation may depend. They work in watches, three airwomen covering the 24 hours. They may be busy or merely sitting and listening. At work they must be calm, quick and accurate; while listening and waiting they must be constantly on the alert, ready to detect the first sound of a signal coming over. They must be intelligent, able to concentrate for long periods, and extremely patient. They must have a thorough knowledge of Morse, and know all the code-sime ond procedure signals connected with their par-

signs and procedure-signals connected with their par-ticular Conjmand. They may be sending or receiving messages between Command and Group, or between aircraft on operations and the base at home. Their integrity and discipling must be beyond custom integrity and discipline must be beyond question.



HAVE received many interesting letters on the subject of musical frequencies which I raised last month. Mr. Ernest Vass, of Aldershot, points out that month. Ar. Ernest vass, of Aldersnot, points out that the earliest known scales were tetrachords, groups of four notes in diatonic succession. St. Ambrose, Archbishop of Milan (374-397), arranged the four diatonic scales known as authentic modes. Later on, Gregory the Great (390-604) added the plagal modes, commoning a fourth below the authentic in the following commencing a fourth below the authentic in the following order:

Authentic	Plagal
Dorian DEFGABCD	Hype-dorian ABCDEFGA
Phrygian EFGABCDE	Hypo-phrygian
	BCDEFGAB
Lvdian FGABCDEF	Hypo-lydian CDEFGABC
Mixo-lydian GABCDEFG	Hypo-mixo-lydian
	DEFCARCD

It would seem that the hypo-lydian gave rise to the modern major scale, and the hypo-dorian to the modern minor scale. Further, the hypo-mixo-lydian duplicated the dorian.

Most of my correspondents seem to agree that the haphazard arrangement of the frequencies were for asthetical reasons rather than mathematical. The harmonic series produces sound pleasant to the ear, although some argue in favour of quarter tones and oriental inusic.

Taking my own selection of intervals, and figures, z to octave, 3 to 2 fifth, 4 to 3 fourth, 5 to 4 major third, 6 to 5 minor third, and 5 to 3 major sixth, there seems a reasonable answer to my question concerning half-tones throughout the octave ; because as there is only a minor third between 5 and 6 one of the intervals must be a semi-tone. We could only have another semi-tone between E and F if the interval between 5 and 6 were a major third. On the other hand, as there is a major third between 4 and 5, we can insert semi-tones between the two steps of a tone which make up that interval.

It is true that by a complete revision of notation, and a complete revolution in musical instrument manufacture, there could be more notes to the octave, and composers would devise harmonious effects, but there are practical implications to be considered. For example, the size of the pianoforte if it were to cover the same compass as at present, the much wider expansion of the figures for octaves, arpeggi, and devising, to say nothing of reading of the necessary expanded form of staff notation.

It emerges from the mass of correspondence I have received that musicians responsible for music are nonmathematical, non-scientific, and that we have arrived at the present stage in music by haphazard and unscientific means. It is too late now to make a change, for it re-writing the world's music. In any case, the present system works, and there seems no reason to alter it.

The New Brains Trust

THE expanded Brains Trust does not seen any more successful than its predecessor, for it has been floored on two or three occasions by comparatively simple questions. The B.B.C. states that the Brains Trust is its most popular feature. Perhaps it is because we all remember our schoolboy days, and we all know the keen delight it gave us to propound questions which the master could not answer. Thus, the item is for entertainment rather than for education. The Brains Trust is intended as a debating society, but the propounders of questions regard it as a baiting society!

British Music versus Jazz I HAVE received the following interesting letter from K. T. Hardman, of Birkenhead, which speaks for itself :

" I think there can be no dispute that you deserve the thanks of listeners as the instigator of the campaign against slush. And most of us hope you will continue in-for we know by experience what the B.B.C. can do in the way of back-sliding. "But I think you are a bit too severe with English

composers and lyric writers, because it is practically impossible for them to get any hearing by reason of the various vested interests they have to contend with. And the worst culprit in this direction is the B.B.C. tself.

" It has definitely educated listeners to swallow the floods of sickly sentimental slush sent over from Tin-Pan Alley-all ready-made and being the least troublesome method of filling in programme time.

"American films also are greatly to blame for the encouragement of such depraved taste—many of them feature a number of alleged 'lyrics' set to whining slush tunes, and thus getting their first 'British boost,' which is later exploited by the song publishers, band leaders, and the B.B.C. "And as there is big money in this vicious circle for

many interested parties, it can easily be understood why any native talent has very small chance of recogis a subject of the second sec between composers and lyric writers, and is always on

tap. "In Great Britain a man or woman may be able to compose most pleasing and catchy melodies, and quite incapable of writing suitable lyrics to fit them to, or vice versa, and it is futile to submit either without the other to any song publisher. Even when a suitable combination between a composer and lyric writer exists or, in rarer cases, where both arc combined in the one person, there is seldom (or never) the cash available for boosting or advertising the native product. And, in consequence, the British publisher will not look at the work, much less publish it." "All of which brings me back to my first point—

that the B.B.C. is the worst culprit and the least anxious for any change. It is cluttered up with too many people who would feel the draught too badly. They are much too well dug in and all of them experts at scratching each others' backs.

"I am no admirer of Lord Haw-Haw, but think he has justification for his remarks about the 'Jazz-soaked British,' and if the general press would follow PRACTICAL WIRLIES' lead, it might give us some hopes of real and continued improvement. We have been so educated with rubbishy filth by the B.B.C. that if Gilbert and Sullivan were reincarnated they wouldn't have a dog's chance to-day."



P.W. Blueprint Service

This Month We are Reprinting the Essential Details of Another Design of "P.W." Receiver

in Popular Demand

(Continued from page 519, November issue.)

Fury Four Super (P.W. 34 C.)

THE circuit arrangement is particularly interesting, embodying, as it does, a number of unusual and particularly good features. For example, instead of using the same method of coupling between the first and second variable-mu valves as between the second variable-mu and the detector, a tuned-grid circuit is made use of in the first position and tuned anode is used in the second. This will at first seem rather unusual, but it has the extremely beneficial effect of rendering the receiver absolutely stable under all conditions, despite the enormous degree of H.F. amplification which is provided. This does not affect the correct ganging of the two-gang condenser, as might be expected, since the coils themselves are accurately matched and are perfectly suitable for this circuit arrangement. In order to ensure that the voltage applied to the screening grid of the two H.F. valves shall be perfectly accurate, a potentiorneter device has been incorporated in preference to the usual series resistance arrangement, and this is in turn provided with decoupling circuits to avoid any possible instability in this part of the receiver. The usual drawback to this arrangement, namely, the continual drain on the hightension battery caused by the potentiometer across the total supply, has been avoided by the method of utilising a four-point on-off switch.

Low Detector Grid-circuit Damping

The detector grid circuit is of rather unusual design, and avoids one of the principal difficulties experienced when high-efficiency coils are employed. The grid condenser, it will be seen, is connected direct to the anode of the second high-frequency valve, and this is in turn connected to a transfer tapping on the grid coil, thus, providing a constant load on long or medium waves and enabling full use to be made of the efficient coil on both wavebands. The grid

circuit is provided also with a change-over switch for gramophone-record reproduction, and the only slight departure from usual practice here is the method of connecting the grid leak to the grid direct, instead of to the grid condenser. Although this means that the leak is in circuit even when the gramophone pick-up is being employed, it enables the leak to be connected in the most direct manner to the grid, and ensures stability on radio by dispensing with an otherwise long lead. The remaining lead, as well as that for one of the pickup connections, is carried in the usual metal braiding, which is earthed, and it will be found that the set is thus perfectly stable on both settings of the switch.

Uniform Reaction Control

The reaction circuit, although perfectly normal as a whole, incorporates a small resistance in series with the condenser in order to prevent any possibility of parasitic oscillation on either waveband, and the small anode by pass condenser forms an essential part of this circuit.

L.F. Coupling

Coupling between detector and output valve is carried out by a parallel fed L.F. coupling device, which includes in the moulded case the anode coupling resistance, a coupling condenser, an L.F. transformer, and yet a further resistance of high value, which is connected in series with the grid of the output valve, as a further precaution against instability. As the specified L.F. Coupling Unit night not now be obtainable, an ordinary L.F. transformer can be used in conjunction with a separate resistor of 50,000 ohms, and a fixed condenser of 0.r.mfd. These should be connected as shown in, the theoretical diagram. Thus it will be seen that at every part of the circuit great care has been taken to ensure that no possibility of instability can occur. The highefficiency pentode valve feeds the loud-speaker through an output filter comprising a tapped choke and large-

(Continued on page 16.)



C1-.0003 mfd. pre-set; C2 and C8-.0005 mfd. ganged; C3 and C12, I mfd.; C4, C6, C7 and C9-.1 mfd.; C5-.0001 mfd.; C10-.0002 mfd. reaction; C11 and C14-.0002 mfd.; C13-.0005 mfd. variable; C16 and C17-.2 mfd.; R1, R5 and R6-50.000 ohms; R2, R3, R7, R9 and R12-1,000 ohms; R4-25,000 ohms; R8-10,000 ohm potentiometer; R10-500 ohms; R11-1 megohm; C15, C18, R13, R14 and R15 are all included in the Graham-Farish Booster Unit.

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capacity fixed condenser, whilst also connected in this part of the circuit is a battery economiser. The purpose of this is to reduce the total anode current required by the pentode when no signal, or a very quiet signal, is being received, and yet permits the current to rise to any required value to produce loud passages without distortion.

The Construction

The first thing to do is to place all the components on the chassis, as shown in the blueprint, and to mark their positions. The screw holes may then be started and the wiring holes drilled. The template which is supplied by the condenser manufacturers should be positioned on the chassis and the necessary clearance holes drilled to accommodate the holding-down screws. When this has been done, the two component brackets should be mounted on the side runners (Fig. 13), and it will be found easier, although not essential, to mount the fixed condensers on these runners whilst they are still unscrewed from the base of the chassis. In this manner the screws are driven home without difficulty.

The Coils

The specified coils will be found to have soldering lugs projecting below the baseplate, and before the assembly is mounted oin. lengths of connecting wire should be soldered to the required lugs as shown by the wiring plan. It will be seen that two of the wires are kept on the upper side of the base and connected to the fixed sections of the two-gang condenser, whilst the remainder pass down through holes in the base for subsequent connection on the underside of the chassis. These holes are shown in the centre of the wiring plan, and the lugs to which they are connected are indicated by the numbers at the side of these holes. The coils are supplied without a switch on the end of the spindle and it is most important that the correct type be used in this position. The switch is a Bulgin Type 87a. The switch must, of course, be mounted so that the small finger at the end of the coil-switch rod operates the dolly of the switch in the correct manner, and this will be when the two end contacts are towards the right, the down the coils, passing the connecting wires down through the holes in the base, and proceed to mount the remainder of the components, using as your guide the wiring plan. There are no difficult points to explain to get at, and there are no pitfalls for the unwary, The receiver is ready for wiring, and this should be carried out in a neat manner, avoiding slackness and making quite certain that good contact is provided at the required points.

Earth Returns

It will be noticed that there are one or two points on the wiring plan marked "M.B.," and these are

earth return connections which are connected to the metallised surface of the chassis, and in the case of those connections on the upper surface it will be sufficient to twist the wire into a loop to be held down by an ordinary round-head wood screw. Where these letters appear on the under-surface it will be necessary to pass a small bolt through the chassis, with a washer be-neath the head making good connection on the upper surface, and the wire held beneath a nut on the under side.

The screened leads must be passed through the metal braiding, and this should be turned back at each end to avoid short-circuits on the inner wire, and small metal brackets used to clamp the braiding to the metal surface. These brackets may easily be made from odd scraps of metal. There is no necessity to screen any

leads other than the two indicated in the wiring plan.

If the specified coils cannot be obtained, we would suggest using three modern screened dual-range coils, accurately matched, and modifying the wiring to suit.

Unscreened coils should not be used in the layout as shown on the blueprint, as instability would, un-doubtedly, result. Some of our advertisers might be able to supply a suitable alternative coil unit.

ORIGINAL LIST OF COMPONENTS FOR THE 1934 FURY FOUR SUPER

One set "Ferrocart" type "G" coils (G.10, G.14, G.13) (with switch—see text) (Colvern). One "Nugang" single variable condenser, .0005 mfd. with type A drive (C.13) (Jackson Bros.). One "Nugang" 2-gang variable condenser, .0005 mfd. with type A drive (C.2, C8) (Jackson Bros.). One disc type H.F. choke (H.F.C.2) (Lissen). One I megohn resistance with wire ends (R11) (Lissen). One nerse aerial condenser. .0003 mfd. (C.1) (Lissen).

- One 1 megohm resistance with wire ends (R11) (Lissen). One pre-set aerial condenser, .0003 mfd. (C.1) (Lissen). One "Pentode" michoke (Varley). One graded volume control, type C.P.158 (R.8) (Varley). One super H.F. choke, type H.F.4 (H.F.C.1) (Bulgin). One fuse holder, type F.5 (Bulgin). One G.B. bias clip, type 2 (Bulgin). Three 50,000 ohm 1½ watt "Ohmite" resistances (R1, R5, R6) (Graham Farish). Five 1,000 ohm ditto (R2, R3, R7, R9, R12) (Graham Farish). One 25.000 ohm ditto (R4) (Graham Farish)

- One 25,000 ohm ditto (R4) (Graham Farish). One 500 ohm ditto (R10) (Graham Farish). One 0002 mfd. reaction condenser (C10) (Graham Farish).
- Tarisn). One "booster" unit (Graham Farish). Two 1 mfd. fixed condensers, type 9200 B.S. (C3, C12) (Dubilier).

- (Dubilier). Four J mid. ditto (C4, C6, C7, C9) (Dubilier). Two 2 mfd. ditto (C16, C17) (Dubilier). One .0001 mfd. ditto type 670 (C3) (Dubilier). Fwo .0002 mfd. ditto, type 670 (C11, C14) (Dubilier). Three 4-pin chassis type valveholders (Clix). One 5-pin ditto (Clix). Four wander plugs marked G.B.I, G.B.2, G.B.3, G.B.+ (Clix). (Clix).
- One Passfeeda coupling unit (B.R.G.). Two large component brackets (B.R.G.).

- Three terminal mounts (Belling-Lee). Oge: 4-way battery cord (Belling-Lee). Six type B terminals (aerial, earth, L.S.+, L.S.--, pick-up,

- Six type B terminals (aerial, earth, L.S.+, L.S.--, pick-up, pick-up) (Belling-Lee). One "Westector" type W.4 (Westinghouse). One "Metaplex" chassis (Peto-Scott). One "Fury Super" cabinet (Peto-Scott). Four valves, types P.M.12M., P.M.12M., P.M.2DX., P.M.22 (Mullard).

r.w.22 (Mullard), One moving coil loud-speaker, type P.M.6 (W.B.). One 120-volt H.T. battery (Siemens), One 16-volt G.B. battery (Siemens). One 2-volt L.T. battery (block batteries). Connecting wire, length metal braiding, screws, etc.



Fig. 13 .- Front view of the chassis of the Fury Four Super.

A Simple Ohm-meter

A Useful Instrument Which Utilises a Low-reading Voltmeter

THE detection of faults in a commercial mainsoperated set is not beyond the capacity of the average wireless enthusiast, and in these days

average wircless enthusiast, and in these days when servicing is difficult to obtain, is sometimes the only way of keeping a set in use.

For this purpose, however, some means of measuring resistances is essential. While an accurate ohm-meter is a comparatively expensive instrument, one that will give the values of resistances to a degree of approximation quite sufficient for most testing purposes can easily be made.

The materials required are an ordinary voltmeter reading to 6 or 10 volts, and a dry battery. A 9volt tapped grid bias battery will be suitable. Connect the positive terminal of the battery to the positive of the voltmeter, as shown in the diagram, and take the negative lead from a tapping that will give nearly fullscale deflection on the voltmeter when the leads are short-circuited—i.e., 9 yolts for a 10-volt meter, or 6 for a 6-volt meter.

Calibration.

To calibrate the instrument it is necessary to know the internal resistance of the voltmeter. In the case of a good-class instrument this will probably be given; or, alternatively, the current taken in milliamperes for full scale deflection may be stated. The resistance can then be found by multiplying the full-scale voltage by r,ooo and dividing by the number of milliamps. Thus a voltmeter which takes 3 milliamps for a full-scale deflection of 6 volts has a resistance of $6,000 \div 3 =$ 2,000 ohms.

2,000 ohms. If the resistance of the voltmeter is not known, it can easily be found, by applying the leads of the ohmmeter to a known resistance. If this resistance has a value of R ohms, and if the meter reads V volts with the leads short-circuited, and v volts in series with the resistance, then the internal resistance of the meter is RV/(V-v) ohms. If, for example, the known resistance is of 600 ohms, and the meter reads 9 volts short circuited, and 6 vo ts across the resistance, then the internal resistance of the meter is $(600 \times 6) \div (9-6) = 1,200$ ohms. The resistance of a cheap voltmeter may be very much less than this. If it is too low the accuracy of the ohm-meter will be impaired by the excessive current taken from the small cells of a grid bias battery. A table can now be prepared showing the voltage readings corresponding to the required range of resistneces. If r is the internal resistance of the meter, V

A table can now be prepared showing the voltage readings corresponding to the required range of resistances. If r is the internal resistance of the meter, V the short-circuited reading, and v the reading with an external resistance R in circuit, then the value of v is obtained from the relation v = Vr/(R + r). Taking r to be 1,200 ohms, and V to be 9 volts, a table such as the following will result.

Ohms	Volts.	Ohms	Volts	Ohr	ns Volts
50000	0.2	3000	2.6	600	6.0
20000	0.5	2000	3.4	400	6.7
10000	· I.O	1500	4.0	200	
5000	1.7	1000	4.9	. 100	. 8.3
4000	2.1	800	. 5.4		9.0

À zero reading indicates an infinite resistance ; i.e., the resistance under test has either broken down or is of a value well outside the range of the instrument, e.g., r or 2 megohns. The table having been propared, a curved strip of paper may be pasted on the face of the meter, and resistance values marked on it against the appropriate voltage readings, as shown in the diagram.

Higher resistances may be measured by substituting a high tension battery for the grid bias battery, and multiplying the observed value of the resistance by the appropriate factor. Thus if a 90-volt battery is available, readings will be multiplied by 10, and on the scale shown a reading of 0.5 volt (20,000 ohms) will now indicate 200,000 ohms. The high tension battery must of course not be used until it is certain from a preliminary test that the resistance is of high value, otherwise the voltmeter may be seriously damaged by excessive voltage across it.

Testing Resistances

When testing resistances in position on the chassis (power, of course, being disconnected) it is necessary to



A voltmeter with its scale calibrated for direct resistance reading.

make sure that there is no parallel path for the current. If a resistance which should be of the order of 5,000 ohms appears on test to be only 50 ohms, it should not be assumed without further examination that the resistance is faulty. It is much more likely that some component in parallel is responsible for the low reading. If the fault cannot be found elsewhere, and the resistance is definitely suspected, it must be removed from the set and tested.

A resistance is often found in parallel with a condenser. It is not necessary to disconnect this. The presence of the condenser, if of sufficiently large capacity, will be indicated by a momentary kick of the voltmeter needle, above the steady value to which it afterwards settles. Provided that no other direct current path is in parallel with the resistance the steady position of the needle will indicate the value of the resistance.

When for any reason the chassis is being examined, it is worth while to make a note of the values of all accessible resistances. Then, should trouble develop later on, a change in the value of any of the resistances will be quickly detected.

fading, are due to the very item which reduces the fading, namely, the A.V.C. action. On the surface, it seems rather a hopeless case; there is, however, one way, at least, of overcoming the trouble to the extent

of making reproduction worth listening to, and this consists of receiving the programme on two different wavelengths and combining their rectified outputs

The whole scheme is based on the assumption that if

both transmissions are subjected to fading, one will

be at its minimum strength while the other is at its peak, thus allowing the combined signal to remain reasonably

constant. It is highly probable that most readers have heard one of the finest examples of "diversity" reception, when they have listened to B.B.C. relays of American transmissions. When these are compared with some of

their earlier efforts, the progress and efficiency of the system is most noticeable. It is very interesting to make a comparison on the spot, so to speak, by receiving the

from the two detector stages.

Diversity Receivers

A Simple System of Reception Which Overcomes Fading. A Fine Subject for Experimental Work

L ITTLE mention has been made recently of "diversity receivers" and the large field of exploration they offer.

Prior to September, 1939, a certain interest was shown in "diversity reception," but since that date little has been published in these pages, except for the description of a suitable H.F. and detector unit, which appeared in May, 1940. The circuit described was designed for medium wave reception, and for the benefit of those interested in such an arrangement, it is reprinted here as Fig. 1. Before discussing it, a simple explanation of the scheme would be helpful to those not familiar with such circuits or their operation. A dictionary definition of "diverse" is various; varied; different, and if we think of these meanings applied to the reception of radio "becomes self-explanatory. A "diversity receiver" is simply a receiver capable of receiving various or different transmissions at the same time. This is not intended to mean various programmes so far as their composition is concerned, but the same programme when it is radiated on different wavelengths. The circuit shown in Fig. 1 is intended to handle two transmissions

-each on a different wavelength -and that is the usual number for simple arrangement. To give an actual example, the B.B.C. Home programme is radiated on wavelengths of 203.5 metres, 391.1 metres, 449:1 metres and 49.34 metres. The pro-Forces gramme is also sent out on three different wavelengths, therefore, the circuit (Fig. 1) could be used to receive one of these programmes on two different wavelengths at the same time. The object of this duplication of the programme is to overcome or reduce certain defects which are



Coils (Bulgin).—C3,4,7 and 8; 0005 mfds. C1, 2, 9, 10, 11, 12 and 16, 0001 mfd. C5, 6 and 17, 01 mfd. C14 and 15, 02 mfd. C18 and 19, 0003 mfd. C13, 8 mfd. R1, 500 ohms. R2 and 3, 0.25 megohm. R4 and 5, 50,000 ohms. R6, 10,000 ohms.

often experienced in the reception of some transmissions. To keep to the local example, i.e., the B.B.C. radiations, in many reception areas the signal strength varies over quite a wide range or, in other words, "fading" interferes with the reception to the extent that the speaker will be handling a good output one moment, and the next, hardly be audible. With a straight circuit, the trouble usually ends there, but with, say, a modern superhet incorporating A.V.C., the fading is accompanied by violent distortion and blasting. The fading itself is not, of course, anything like so pronounced as with a straight set—owing to the A.V.C. action, but the other two objectionable forms of interference, and these can be more annoying in the writer's opinion than pure here. From the above, it will be obvious that "diversity reception" can be employed for the short-waves, for the reception of those stations which radiate the same programme on two or more wavelengths.

Circuits

The present time may not seem an opportune moment to suggest circuits using more valves and components than those normally required; therefore, for experimental purposes it is suggested that the simplest arrangements are used. In those districts where, for example, the reception of the Home or Forces programmes are marred by blasting and distortion consequent upon fading, and where a multi-valve superhet is in use, it would be a proposition to consider constructing a receiver similar to that shown in Fig. 1, and feeding its output into the L.F. side of the superhet. The valves preceding the L.F. portion could be turned to good account in the "diversity reception" section, the latter getting its H.T. and L.T. from the mains unit in the set.

For battery operation, the circuit would be fundamentally the same as Fig. 1, ignoring, of course, the cathode and bias circuits and using the usual arrangements common to battery sets.

A simple two-valver is shown in Fig. 2. It consists of two detector stages, each with its own aerial and reaction circuits. In the anode circuits, a normal resistancecapacity coupling in each case, the output sides of the two coupling condensers being made common to each other, and thus forming one side of the output from the two stages. The other side is taken from the commonnegative-earth line.

The aerial coils required will depend on whether the set is to be used on medium or short-waves. If the former, the medium-wave section of any good dual-range coil, with reaction winding, can be used, or a simple coil made to one of the specification's which have appeared in past issues. For short-waves, the ordinary 4- or 6-pin coil can be used, and it is the latter which is shown in the diagram.

The aerial—common to both stages—is fed into the aerial coils through variable condensers (fixed can be used) of, say, ocor mfd. maximum capacity. The remainder of each stage is perfectly orthodox.

The output from the two valves can be fed straight on to the grid-of and output pentode or the first valve of a two-valve L.F. amplifier, provided it has a grid resistor connected to it and can obtain its required bias. For test purposes, a pair of phones can be connected direct to the output terminal and the earth line.

Operation

There is little that can be said about the operation of such an arrangement, except for a tip about the tuning of each stage. To use the B.B.C. transmissions again for an example, we will assume that the programme required is the Home one, radiating on three mediumvaves and one short. We will also assume that the circuits are designed for the medium-wave reception:



Fig. 2.—For S.W. diversity reception experiments this detector arrangement could be used.

the procedure is, remove valve 2 and on circuit r tune in, say, the 203.5 metre transmission, making the necessary adjustments of the reaction circuit, etc., in the normal manner. When satisfied, remove valve r and insert valve 2 in circuit number two, and on that circuit tune in the same programme on the 449.r metre wavelength, or the 391.r metre, if that gives better reception. Make all final adjustments as before, and then valve r can be plugged back into circuit 1, thus bringing both valves into action. The output should now be satisfactory and hold reasonably constant, provided care has been taken to ensure that neither of the circuit is oscillating or on the verge of oscillation, and that the individual outputs are, as near as one can judge by ear, the same as regards volume.

Notes from an Amateur's Log-book

Which is the Best "Straight" S.W. Circuit? 2CHW Gives His Views Based on Recent Experiments

HAVE been making tests with 1-v-o and o-v-I arrangements to determine the relative advantages of H.F. and L.F. amplification.

The subject is one which always serves as good material for debate when amateurs get together. There are three schools of thought; one which maintains that an 'H.F. stage is essential, another which takes an intermediate view and claims that it—an H.F. stage—is an asset, but only roo per cert. efficient below a certain frequiency. The third party maintains that a good detector followed by L.F. amplification to suit conditions, is the best all-round arrangement in the "straight" class of S.W receiver. There seems little hope that the three schools will become reconciled to a single view or theory, which is a good thing, as each one can give practical proof to support its claims. The tests I carried out were purely aural, and not in

The tests I carried out were purely aural, and not in any way along the lines one would take if laboratory facilities were available. However, they were interesting and I would suggest that the subject is one which could be examined more thoroughly by those S.W. enthusiasts who wish to carry out some experimental work during the long black-out evenings.

The apparatus required can be the Rx in use, if it has

the following valve sequence, I-v-I, or if the circuit is a simple v-I, then the necessary additional components, will no doubt be available from most "spares" boxes My work was simplified considerably by the design of the Rx in use, the theoretical circuit of which is shown in the diagram. Those who have followed my previous notes will remember that the H.F. stage included switching which enabled it to be used tuned or unfuned. It is also a simple matter to transfer the aerial to the detector, one or two stages of L.F. can be used as desired, therefore the complete circuit lends itself to practically any combination.

Observations

To secure the most rapid comparison during any of the tests it is necessary to provide switching to enable the required circuit changes to be made quickly, otherwise, it is possible for very confusing results to be obtained, fluctuating conditions which might be present during the testing period. Use switches having good contacts—especially on the H.F. side—and keep wiring reasonably short.

Prior to the actual tests, I had a general search

around the dial to see how conditions were and soon found that they could be classified as "fair to good." The W's seemed to be the best transmissions to work on, so with the circuit reduced to r.v.t I made a careful note of their reception, giving each transmission a reasonable period so that its average, so to speak, could be obtained.

Switching over the H.F. stage to "untuned" and readjusting the "spreader" across the detector tuned circuit, I found that the difference in signal strength was not worth noting, though there did seem to be a greater likelihood of interference. The reaction control could be increased slightly and this tended to balance matters. The tests were carried out on the 16 m., 19 m., 25 m. and 30 m. bands.

Going up on to the.45 m. band, the tuned H.F. stage offered an appreciable improvement over the untuned, and this becomes more marked as the wavelength is increased, therefore the results certainly supported the theory that there is little or no advantage to be obtained by tuning the H.F. stage on transmissions below, say, 40 metres.

1-v-0 versus 1-v-2

The signal-to-noise ratio, using the I-v-I arrangement,

'Dead spots," loss of reaction or erratic operation, poor selectivity and attendent problems are likely to be experienced; in fact, I only eliminated, or should I say reduced, the first two by careful and frequent adjustments of the aerial coupling coil and a 50 nmfd. variable condenser I tried in series with the aerial. Fortunately, the aerial in use is an ordinary inverted "L" type, and not designed or cut to resonate on any one particular wave-band; therefore, I did not have too much trouble. Many amateurs, thinking to improve matters, will often take the trouble to erect a tuned aerial system. This is all right up to a point, but it must not be overlooked that at resonance, or harmonics, the aerial will impose a heavy load on the detector circuit, a load often sufficient to stop the circuit oscillating. Given reasonable aerial and earth systems, and provided care is taken in "hottingup" the circuit, there seems every justification for the claim of the o-v-r school, that such a set can pull in the DX transmissions.

Bringing in the other L.F. stage gave excessive volume, and spoilt the signal-to-noise ratio. While admitting that there are times when one wants just that little extra amplification, it seems that if two L.F. stages are to be included, inter-valve switching and/or some form of volume control must be incorporated.





was quite good, so the extra L.F. stage was brought into circuit just to note its effect. The signal strength was increased beyond comfortable value, but the background noise came up to a greater extent than with the one L.F., thus giving a worse signal-to-noise ratio.

Going to the other extreme, I then tried r-v-o, i.e., just the H.F. and detector stages. Signal strength dropped, noticeably, of course, but the ratio—S to N was as good as one could wish, even if I did have to concentrate a little more on the actual listening. The combination was beautiful to work; reaction was as smooth as velvet, and the signal seemed to "stand out." This is not really surprising when one remembers what can be done with a single detector valve in the hands of an experienced enthusiast, and when it is used in conjunction with a good aerial. The results almost convince one that it would be worth while scrapping the L.F. stages and concentrating on the perfecting of H.F. amplification and the aerial. Whatever views are taken, it cannot be denied that the idea offers plenty of scope for winter experiments. Although it will appear to be a contradiction, during the above test I would say that the best results were obtained with the H.F. stage tuned.

0-v-1 and 0-v-2

The simple, well-tried o-v-I arrangement gave quite good results, but its defects were obvious when rapid comparison was made with the I-v-o and I-v-I circuits. The signal strength and signal-to-noise ratio was, speaking generally, satisfactory, but it soon became evident that the loading of the aerial directly on the grid circuit introduced some of those defects which can become the bugbear of the amateur's activities.

Summing Up

From these simple tests one thing stands out most prominently, namely, the real value of an H.F. stage, whether it is tuned or untuned. Even if little or no amplification is obtained below 40 metres the stage serves the very useful purpose of "buffer" between the aerial load and the detector grid circuit, thus reducing, if not eliminating, "dead-spots," unsatisfactory reaction control, loss of sensitivity and poor selectivity. An arrangement of the 1-v-1 type seems to be the best in the "straight" class of S.W. receiver.

A I-v-O combination has possibilities well worth considerable attention from those interested in DX'ing at headphone strength.

The o-v-I circuit is best classified as a very good line-up when in the hands of an experienced enthusiast and operated in conjunction with a good aerial and earth.

More depends on the overall efficiency of the station or installation—and this includes the skill of the operator than many amateurs think. It is not unknown for a really good Rx, in some cases a reliable commercial product, to be condemned by an owner as being far below par, whereas, in actual fact, the fault quite possibly is in the operator and/or the aerial and earth system. This brings forth the justified argument that all S.W. enthusiasts should serve a thorough apprenticeship with Rx's of the o-v-o and o-v-r types. With such simple receivers, one has, of necessity, to acquire a reasonably high degree of skill and touch; perfect acrial and earth arrangements and a knowledge about local and atmospheric conditions with relation to the reception of DX transmissions. December, 1942

Practical Hints

An Accumulator Support

Lugs soldered

to frame

FTER having an accumulator A tipped over in my radio set I have fitted the supporting device shown in the sketch. The materials required are :

Two pieces of hard wood 1in.×1in. thick. The height of the stand is made to suit accumulator, and allowing 12in. for clearance at the bottom.

Two pieces of brass rod lin. diameter and rin. long are soldered to a strip brass frame, which fits the

THAT DODGE OF YOURS!

THAT DODGE OF YOURS! Every Reader of "PRACTICAL WIRE-LESS" must have originated some little dodge which would interest other readers. Why not pass it on to us? We pay 3:10-0 for the best hint submitted, and for every other item published on this page, we will pay halt-a-guinea. Turn that idea of yours to account by sending it in to us addressed to the Bditor, "PRACTICAL WIRELESS," George Newnes, Ltd., Twee House, South-'ampton Streed, Strand, W.C.2. Put your name and address on every item. Please note that every notion sent in must be originad. Mark envelopes' "Prociola Hints." DO NOT enclose Gaeries with your hints.

SPECIAL NOTICE

All hints must be accompanied by the coupon cut from page iii of cover.

A Radiogram Conversion

RECENTLY had an old washing stand given to me for use as a bench, but I hit upon the following idea for converting it into a radiogram.

I removed the marble top and door, and replaced the that the top with a piece of §in. plywood. I next fitted a 3/16in. plywood panel where the door used to be. I then took my 4-valve com-mercial mains receiver chassis and mercial mains receiver chassis and loudspeaker out of their original casing and fixed them to the inside of the washing stand by means of nuts and bolts, after which I took my clockwork motor and turntable, etc., and put them in the washstand, in the

Electric Lamp

position shown. With a short piece of screened twin flex I connected the pick-up to the pick-up terminals, and, of course, connected my mains and A.E. leads to their appropriate terminals.

Cigarette Lighter



A useful dodge for preventing the spilling of an accumulator.

accumulator like a carrier. Slots are cut in the top ends A. A. HEAD (Writtle)

Improvised Tinplate Chassis

A the present time there is a great shortage of aluminium and cheet iron with which to make chassis for wireless sets. I have found that a good consist for whereas sets. I have found that a good substitute is an upturned tinplate baking tin, the components being mounted on it, as shown in the accompanying sketch. This forms an effective screen for a work bench job, and will, of course, take solder easily.—J. M. FIRTH (Beckenham).





How an old washing stand was converted into a radiogram.

I found that the P.U. was a little in the dark, so I connected a 270v. 30w. 221mm. Striplite and switch in position. (See sketch.)

As a refinement, I fitted a novel electric cigarette lighter, which can be used only when the set is in operation.

I took a double-pole tapped filament valve-holder to use with one of the valves in varye-noncer to use with one of the valves in the set, and connected it to a 5 amp. 2-pin plug, and across the two pins I connected a piece of nickel chrome clement wire 26 S.W.G., about 4in. long, and wound into a small coil. The lighter is placed on top of the radiogram the deve converting the of the radiogram, the flex connecting lead passing through a hole in the left-hand panel, as indicated in the sketch. — W. WILLIAMS (Bristol).

PRACTICAL WIRELESS

December, 1942



A front view of the completed instrument, with tape in position ready for operation.

THE morse recorder described below should be of great use to all those who are learning the morse code, whether singly or in groups. It can be made almost entirely from material found in the junk box, and the total cost should not be more than about five shillings for the parts that have to be purchased. A lathe is not necessary, and the only tools needed are those found in the average amateur's workshop. The recorder has been tested at speeds up to about 30 w.p.m., but it is estimated that speeds of at least roo w.p.m. can be recorded satisfactorily. The recorder operates in the following manner. The paper tape C (Fig. 1) is pulled through the

The recorder operates in the following manner. The paper tape C (Fig. 1) is pulled through the machine at a constant rate by means of an external tape puller. Fetween the tape and the striker D there passes the typewriter ribbón B, which runs between the typewriter spools AA. These spools are driven by a motor F via the worm gear G and the bevel gears HH, which are mounted on the shaft L and located behind the spools (the gears are clearly shown in Fig. 2). The direction of travel of the ribbón is changed by moving the shaft L laterally and so engaging one or other of the pairs of bevel wheels:

The striker D and the armature M (Fig. 3) is actuated by the electromagnet J, which is energised by an alternating, current (50 c.p.s.) keyed by a suitable relay, or directly from a telegraph key. It was found necessary to use alternating current in order to obtain a clear impression on the tape.

Each of the bobbins are wound with 80 turns of 28 S.W.G. enancellel copper wire, the windings being connected in series and energised by a transformer giving 3 volts R.M.S. It was found that when using D.C. to energise

It was found that when using D.C. to energise the windings, the pressure of the stufficient to stop when making an impression, was sufficient to stop the typewriter ribbon. The result of this was that at slow speeds or on long dashes, the imprint on the tape, although black at the beginning of a dash, speedily became weaker initil it was almost invisible due to the fact that no fresh ribbon was moving under the striker. On the other hand, with A.C. this effect does not occur since the striker is not in contact with the ribbon throughout a dash, thus allowing the ribbon to move. A Morse

The Construction and Operation of a

By D. McM

A clear impression results, however long a dash is made.

Constructional Details

The construction of the instrument, should be clear from the illustrations, and the design can be modified to suit the materials on hand. The base, the electromagnets, and the armature were taken from an old moving-iron loudspeaker, the bobbins being rewound to suit the transformer used for supplying the current. The typewriter ribbon may be obtained from any stationers, and an extra spool should be asked for at the same time. The spools are attached to their shafts in the following manner. Remove the side of the spool and solder a collar on its inner face. A locking screw should be fitted to the collar. The side

is then replaced, the shaft pushed through and the locking screw tightened. Brass gears were used, and the motor, F, was a low

Brass gears were used, and the motor, F, was a low voltage motor of the type used for driving small model boats: any small motor would be suitable. There is one point to note, however, and that is that the motor should not run too fast, otherwise a very short time will elapse before the direction of travel of the ribbon has to be changed. As previously stated, the reversal is made



Fig. 3.—Showing the magnet, armature and striker assembly in position on the bedplate.

by moving the shaft L laterally, and it s

in one position or the other by means of a movable collar held on the shaft by a grub screw. This screw is loosened whenever it is desired to reverse the ribbon, and the other pair of bevel gears engaged. A small lever could be fitted instead of the collar to move and lock the shaft, and in many respects would be more convenient. especially if the motor runs rather fast.

Recorder

of a Useful Unit for the Experimenter

MCMULLAN

The springs N were taken from an old 'phone jack, and these were soldered to the armature M. The rest of the instrument may be constructed from scrap metal-aluminium was actually used in the model illustrated. although brass would do equally well. The whole unit is held together by means of the long screws PP. The end of the striking screw should be filed to a chisel point, but should not be too sharp or it will tear the ribbon.

Paper Tape

The paper tape (in, wide) may be obtained from Webb's Radio, of London. Both sides of the tape may be used and, if the striker is attached in such a way that it can be moved across the tape, two or more impres-sions may be recorded on the same side of the tape. One reel will last a long time, especially if this procedure is adopted.

The tape puller was made from an old gramophone motor by removing the turntable, and slipping a piece jockey wheel was also fitted, and the tape was threaded between it and the rubber-covered spindle. An electric gramophone motor would be best but a clockwork one will answer the purpose.

The instrument may be used for checking the accuracy of one's key work, and for this it has the advantage over the "wavy line" recorder that it records actual dot's and dashes which are more easily read.

Radio Transmission

For recording radio transmissions, those stations not employing modulated C.W. are first heterodyned with a B.F.O. to produce an audible note. The A.F. alternating current is then rectified and the resulting D.C. used to energise the relay. The relay was constructed from an old buzzer and is satisfactory up to speeds of about 30 w.p.m. I have not yet had time to construct a relay suitable for higher speeds, and the present relay does not merit description.

Morse Practice

When undertaking morse practice, it is vitally important to acquire correct timing and spacing, and even if the test signals are being received by another person, it is not always easy to correct taults in these directions.



Fig. 2 .- The shaft "L," bevel gears "HH" and the worm drive are exposed in this view.

When the student is operating alone, it is practically impossible for him to check up on the length of the dots and dashes and the pauses between letters, etc., and words, with the result that he may develop a far from perfect style. A morse recorder, such as that described above, enables one to have a permanent record—in black and white—of the signal as it was keyed, thus allowing faults to be revealed and corrected.



Radio Examination Papers-13

Another Set of Random Questions, with Suitable Answers by THE EXPERIMENTERS

-Tuning with A.V.C.

quite sound practice. But when a good measure of A.V.C. is obtained this

cannot give satisfactory results because the sensi-

tivity of the controlled

stages (or more correctly,

the gain of these stages)

increases as the input to

the A.V.C. valve decreases.

is slightly de-tuned the gain of the controlled

valves rises, and in many

cases the output increases,

so that the volume of reproduction is greater

than when the set is ac-curately on tune. Some-

times this change in output

is very pronounced, in

which case fairly accurate

tuning may be possible by

choosing the point midway between the two " peaks." In general, however, this is

impracticable, possibly be-cause ionc "peak" is more pronounced than the

other.

Thus, when the rece ver

HEN tuning a receiver by aural methods, that is by listening to the speaker output, it is customary to tune for the maximum volume. When A.V.C is not provided, and sometimes when a measure of A.V.C. is available, this is the filaments are treated would be "sucked off" and the filament might even collapse.

The results would be similar, but "in reverse" if the L.T. were switched off before the H.T., since the anode voltage would rise while the filament current was quickly, falling to zero.

QUESTIONS

- 1. Why is it extremely difficult to tune accurately by the aural method a receiver fitted with amplified A.V.C. ?
- 2. In the case of a battery set operated from an H.T. eliminator it is important that the L.T. should be switched on before the H.T. and off after the H.T. State why.
- 3. Explain the principle of the Wheatstone Bridge, and show how it is used in the measurement of resistance.
- 4. What would be the effect on reproduction if the by-pass condenser across the bias resistor of an oulput pentode became open-circuited ?
- 5. What do you understand by the "kink" in the characteristic curve of an S.G. valve?
- 6. If it were required to operate a four-value " all-dry " receiver from a L.T. accumulator how would you make the necessary resistor?

Details of a few gauges of Eureka resistance wire are given in the following table :

	S.W.G.	Current Capacity (amps.)	Ohms/Yard
	24	1.0	1.8
	26	-5	2.6
	28	.25	4.0
,	30	.2	5.7

In such circumstances a

"in step " with the A.V.C. voltage, and shows the proper tuning point when the A.V.C. voltage reaches its peak. And that peak is reached when the tuning circuits are accurately tuned to the required transmission, or to the I.F. in the case of the I.F. stages of a superhet.

It may be argued that aural tuning can be used if one listens for distortion instead of for maximum volume, bearing in mind that side-band distortion occurs unless the tuning is correct. In practice, however, it is almost impossible to use this method effectively.

2.—Eliminator Switching The output voltage of an climinator varies with the current load, being highest when the current is zero. This is especially the case with an A.C; eliminator which incorporates a rectifier. Thus, if the H.T. were switched on before the L.T. it would have a much higher voltage than that at which it was rated. In consequence, the voltage applied to the various smoothing and by-pass condensers would be excessive, and these components may be damaged.

Further, when the L.T. was switched on the anode voltage would be excessive for a fraction of a second during the time the filaments were heating up. As a result, the rare-earth oxide with which



3.--The Wheatstone Bridge The Wheatstone resistance bridge consists of a network of resistors through which a current is passed from a D.C. source of supply. If the resistors are arranged as the sides of a square as shown in Fig. 1, and a galvanometer connected between two diagonally - opposite corners of that square, a battery being connected between the other corners, it is possible to adjust the values of the resistors so that there is no current flow through the galvanometer. This is known as balancing the bridge.

The condition necessary for balance is that the po-tential at the two points to which the meter is connected shall be the same. This means that the ratio between the values RI and R2, and R3 and R3 must be the same, Expressed

differently: RI/R2 must equal R3/Rx, in Fig. 1.

It can be seen that if Rx is the resistor whose resistance it is required to find, the bridge could be balanced by variation of R3.

In practice, a simpler method of balancing is used. A length of resistance wire (generally one metre long and mounted on a metre rule) is used to replace RI and R2, Indifferent of a fixed resistor of the same order as the resistor to be measured, is used in place of R_3 . The practical connec-tions, then become as shown on the right in Fig. 1. It will be seen that although the various parts are shown in

different physical positions, the circuit is exactly the same as before.

There is a slider which moves over the length of resistance wire, and this is moved from end to end until the bridge is balanced; that is, until the centre-reading galvanometer shows a zero reading. As the resistance wire has a uniform resistance per centimetre over its whole length, the two distances from the ends of the wire to the slider can be used in place of the exact resistance (in ohms) of the two parts of the wire. We can then write our equation: $\frac{1}{2}R/Rx$. Since we know the two lengths and also the value of R, we can find the value of the unknown resistance Rx. This is simplified by rewriting our equation as: $Rx = \frac{R \times l_2}{1}$

For example, if lr were 75 cm. and l2 25 cm., while R had a value of 5 ohms, the value of Rx would be 15 ohms.

4.—Open-circuited Condenser

In the first place it would probably be found that there would be a certain amount of mains hum, due to an A.C. potential developing across the resistor. This would not always be in evidence, however, cspecially

if the resistor were of low value. But there would be a reduction in output volume due to the introduction of a form of negative feed-back.

This negative feed back would $\widehat{\mathbf{T}}^{8}$ occur due to the automatic bias across the resistor acting in "opposition" to the bias applied to the grid by the audio-frequency input. In this case, the automatic bias would be varying at audio frequency due 3 5 to the absence of the by-pass condenser.

Suppose the grid were going more negative at any instant under the effect of the audio voltage. This would have the effect of reducing the anode current. But as the anode current fell, so would the current through the bias resistor. That would mean that the voltage across that re-sistor-which is used as automatic bias -would also fall. In other words, the grid would tend to become more positive in respect of the cathode. This would

be equivalent to reducing the amplitude of the A.F. on the grid, so bringing about the reduction in volume.

5.-S.G. Characteristic Curve

A representative anode voltage-anode current characteristic curve for a screen-grid valve is shown by a full line in Fig. 2, where a corresponding curve for a triode valve is indicated by a broken line. In the case of the curve for the S.G. valve, it will be seen that, as the anode voltage begins to rise from zero, the anode current also rises. But after the voltage has reached about 15 (the exact voltage varies with different valves and different operating conditions) the current starts to fall in spite of the continued rise in anode voltage. After the voltage has reached about 30, the current again starts to rise, and continues to do so up to the maximum anode voltage.

The irregularity in the shape of the curve is described as the "kink," for obvious reasons. The cause of the "kink" is the release of secondary electrons by the anode to the screen grid. This is due to the fact that anoue to the screen grid. This is due to the fact that the primary electrons are accelerated by the screening grid, strike the anode with such velocity that they cause other electrons to be knocked off; these are attracted to the screen, which has a higher positive potential. As the anode voltage is increased slightly above a certain critical forms a reacting mean the above a certain critical figure, a negative space charge

is built up between the anode and screening grid, and this resists the return of the electrons to the screen. Once the anode voltage becomes cyual to the screen voltage, the anode can "hold " all the electrons which strike it.

To avoid the effects of the "kink" it is necessary to use an anode voltage at least 50 per cent. higher than the screen voltage.

6.-Limiting Resistor Since "all-dry" valves work at a filament voltage of r.4, it is necessary to use a dropping resistor to bring down the voltage of the 2-volt accumulator from 2.0 to 1.4; thus it is necessary to drop .6 volt. Before we can find the value of a resistance suitable for this we must know the total L.T. current taken by the receiver.

The average modern four-valve battery receiver has a frequency-changer followed by a screen pentode, double-diode-triode and then by a pentode output valve. The first three of these will each have an L.T. consumption of .05 amp., while the output valve will take .r amp. Thus, the total consumption will be .25 amp.



Fig. 2.—The anode current-anode voltage characteristic curve for an S.G. valve is shown by the full line. The broken line shows a corresponding curve for a triode.

It is an easy matter to find the value of resistor required to drop .6 volt at .25 amp. by dividing the voltage by the current. Thus the value (R) of our resistor should be .6/.25, which works out very slightly over 2 ohms. In practice, it would be quite sufficiently accurate to use a value of 2 ohms.

Any of the gauges of Eureka resistance wire shown in the table could be used, but since the 30-gauge wire is rated at only .2 amp. it would probably warm up in use. As changes in temperature affect the resistance, it is therefore better to use wire of one of the other gauges; we could use whichever was more readily obtainable. Thus, we could use $\frac{1}{2}$ yd. of 28-gauge wire, The resistor could most conveniently be made by

winding the wire on a strip of paxolin. As the wire would probably be bare, the turns would have to be slightly spaced, either by making "nicks" in the edges of the paxolin with a file, or by winding the wire with a length of string running alongside it.



December, 1942

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of figures before the decimal point in the divisor, unless the quotient appears to the *left* of the original position of the cursor, in which case the number so determined is increased by 1.

Multiplication or division on the slide rule should always be accompanied by mental addition and subtraction, so that the position of the decimal point in the result is known even before the figures are read from the scale. This applies particularly when a calculation involves a number of such steps, as, for example, in evaluating

 $247 \times 3.3 \times 0.179$ It may be noted that in the case of a decimal the equivalent to the "number of figures before the decimal point" is minus the number of o's between the decimal point and the first figure. For example 0.0031 is classed as - 2.

In dealing with logarithms, the "characteristic" is I less than the number of figures before the decimal point. When using the slide rule, this fact is best forgotten as it may lead to some mental confusion.

Accuracy of Slide Rule

The slide rule is obviously an instrument of limited whereas the corresponding arithmetical accuracy, processes may, if desired, be carried out with perfect accuracy. What must be remembered, however, is that for many engineering calculations the slide rule is accurate enough. In other words, the figures upon which the calculation is based often include unavoidable errors, probably much greater than those introduced by the slide rule, and so its defects are unimportant.

Modification of Formulæ

It sometimes happens that modification of a formula will make it possible for the slide rule to evaluate it in some particular case with much higher accuracy than could be done at first.

Thus, if it be required to evaluate
$$\frac{17.313}{12.702}$$
, direct

division by slide rule gives 1.256, but the last figure 6 is uncertain, and anight be read as 5 or 7. Higher accuracy, with little extra time, can be achieved by writing

$$\frac{17.313}{13.792} = \frac{13.792 + 17.313 - 13.792}{13.792} = 1 + \frac{3.521}{13.792}$$

=1+0.2555=1.2555 and this is correct to ± 0.0002 . The principle involved here is to subtract unity from the given quantity, and then to use the slide rule to evaluate the remaining fraction.

Similarly, the slide rule is used in the ordinary way to show that $\frac{55.314}{3.065}$ = 3.065, but the last figure is uncer-

18.052 tain. This result is equal to about 3, and in order to attain

a more accurate result the given quantity is therefore reduced by 3, thus,

$$\frac{55.314}{18.052} \xrightarrow{3 \times 18.052 + 55.314 - 3 \times 18.052}{18.052}$$

-54.130 -3 - 18.052 =3.0641 and this is $=3+\frac{22\cdot3+4}{2}$ 18.052

correct to \pm 0.0002.

Again $\frac{11.413}{2.912}$ = 3.92 and, as this is a little less than 4_{x} the difference between the original quantity and 4 is. determined thus:

11.413 4×2.912+11.413-4×2.912

$$=4+\frac{11.413-11.648}{2.912}=4-\frac{0.235}{2.912}=4-0.0807=3.9193$$

2.912 1 2.912 and this is correct to ±0.0002.

These methods, it will be noted, include first a quick determination of the result by straightforward slide rule working and then a more accurate determination by a different method. The total time occupied is less than would be required for long division, and, in addition to this, the two-way method affords the valuable safeguard

of an independent check, so that an accidental error is fairly sure to be noticed.

The same principle of using the slide rule to calculate the difference between the desired quantity and some slightly different quantity that can be determined quickly and accurately may be used in other cases. For example, it may be desired to evaluate the expression

 $\sqrt[n]{(a^2+b^2)}$ in a case where b is small compared with a. To adapt this expression to accurate slide rule calculation in such circumstances it is transformed thus

$$\frac{a}{(\overline{a^2 + b^2})} = \sqrt{\left(\frac{1}{1 + \frac{b^2}{a^2}}\right)} = \left(1 + \left(\frac{b}{a}\right)^2\right)^{-1}$$

By use of the binomial theorem, this last quantity is expressed in a series of terms thus:

$$\sqrt{\frac{a}{(a^2+b^2)}} = \mathbf{I} + (-\frac{1}{2}) \left(\frac{b}{a}\right)^2 + \frac{(-\frac{1}{2})(-\frac{3}{2})}{\mathbf{I} \times 2} \left(\frac{b}{a}\right)^4 \dots \text{ etc.}$$

If g is much less than unity, successive terms in this

series become rapidly smaller, and only the first one or two need be considered. If, for example, a = 14.72 and b=2.31, the original expression is

$$\frac{14.72}{4.72^2 + 2.31^2} = \frac{14.72}{222.35} = \frac{14.72}{14.91} = 0.987$$

 $\sqrt{(14.72^2+2.31^2)}$ 222.35 14.91 but there is some doubt about this, because the 14.91 might be 14.90 or 14.92.

Using the modified form of the expression, its value is

$$\mathbf{I} + (-\frac{1}{2}) \frac{(2.3\mathbf{I})^2}{(2.3\mathbf{I})^4} + \frac{(-\frac{1}{2})(-3/2)}{(-3/2)} \frac{(2.3\mathbf{I})^4}{(2.3\mathbf{I})^4} \dots$$
 etc.

=0.9879.

1

This is correct to ±0.0001 for the terms considered, and it is clear from the rapid reduction in passing from the first to the third that all others will be negligible.

Application of the binomial theorem succeeds here because it is used for the expansion of an expression of the form $(1+x)^n$ where x is much smaller than unity. This means that succeeding terms diminish rapidly, so that all beyond the first few are negligible. If this were not the case the method would be useless, as the whole

expansion involves an infinite number of terms. For convenience of reference the expansion of $(1+x)^n$ is given here in general terms. It may be noted that the more general form $(a+x)^n$ is sequal to $an[1+(x/a)]^n$, so that if x/a is small the theorem can still be usefully employed in this way:

$$(1+x)^n = 1 + nx + \frac{n(n-1)x^2}{1 \times 2} + \frac{n(n-1)(n-2)}{1 \times 2 \times 3}x^3 + \text{ etc.}$$

Succeeding terms are easily added. The power of xin any term is I more than that of the preceding term. The numerator of the coefficient includes all the terms of the preceding coefficient and an extra one that is I less than its predecessor. The denominator of the coefficient includes all the terms of the preceding coefficient and an extra one that is I more than its predecessor.

Some common occurring examples of this application of the binomial theorem are given below for convenient reference :

$$\frac{1}{1+x} = 1 - x + x^2 - x^3 + x^4 -, \text{ etc.}$$

$$\sqrt{(1+x)} = 1 + \frac{x}{2} - \frac{x^2}{8} + \frac{x^3}{16} - \frac{5}{128} x^4 +, \text{ etc.}$$

$$\frac{1}{\sqrt{(1+x)}} = 1 - \frac{x}{2} + \frac{3x^2}{8} - \frac{5x^3}{16} + \frac{35}{128} x^4 -, \text{ etc.}$$

$$(1+x)^2 = 1 + 2x + x^2 \text{ (complete).}$$

$$\frac{1}{(1+x)^3} = 1 - 2x + 3x^2 - 4x^3 + 5x^4 -, \text{ etc.}$$

$$(1+x)^{3/2} = 1 + \frac{3x}{2} + \frac{3x^2}{8} - \frac{x^3}{16} - \frac{3x^4}{128} -, \text{ etc.}$$

$$1/(1+x)^{3/2} = 1 - \frac{3x}{2} + \frac{15x^2}{8} - \frac{35x^3}{16} + \frac{315x^4}{128} -, \text{ etc.}$$

Small Differences

The calculation of a quantity by means of a formula that expresses it as a relatively small difference between two large quantities tends to be inaccurate and for that reason an endeavour should always be made to modify The formula in such a way as to minimise this defect. For example, suppose that a quantity e is the difference between the square of two quantities d and e, that is

 $c = d^2 - e^2$ Now in a particular case let d=12.31 and e=11.69: By slide rule

12.31²=151.6±0.2

In this case

Subtracting, c = (151.6+0.2) - (136.6-0.2) = 15.4or c = (151.6-0.2) - (136.6+0.2) = 14.6These results represent the limiting values of c corresponding to the uncertainty of \pm 0.2 in the slide rule reading.

Higher accuracy can, however, be achieved by using the modified formula $c=d^2=e^2=(d-e)(d+e).$

c = (12.31 - 11.69) (12.31 + 11.69)



Examples in Slide-rule Calculations.

By this method, slide rule working is reduced to a simple multiplication and the unavoidable error in reading is confined to the final result. It is only one-

twentieth of the possible error by the obvious method. Similarly such expressions as $(d^3 - \epsilon^3)$, $(d^4 - \epsilon^4)$ and so on, may be transformed to give higher accuracy in slide-rule computation, and for convenience, a number of the more commonly occurring cases are tabulated below:

Trigonometrical Quantities

COS

The slide rule is not always used as freely as it might be for evaluating trigonometrical functions of angles. This is largely the result of lack of practice, but once that has been overcome, and confidence attained, the slide rule within its limits of accuracy is much quicker and easier to use than are trigonometrical tables.

It has some limitations, but they can be overcome It has some limitations, but they can be overcome without much trouble. For example, it does not show the tangent of any angle less than 5° 43', but it does show the sines of angles down to 0° 34', and for angles between that and 5° 43' the tangent may, for most purposes, be regarded as equal to the sine. More accurately, for angles less than 5° 43'

$$tan A = sin A + \frac{sin^3 A}{2}$$

For example, the slide rule shows that $\sin 4^\circ = 0.0695$ and therefore, using the formula given above,

$$tan A = 0.0695 + \frac{(0.0695)}{6}$$

=0.0695 + 0.00006 = 0.06956.

It will be realised, however, that the last figure in 0.0695 (read as the sine of 4°) cannot be guaranteed, so that this correction is a fine one. For all angles less than 0° 34' the sine and tangent

may be taken as equal to the number of minutes divided by 3440.

Another range in which the slide rule fails to give accurate direct readings is for cosines of small angles. This defect is overcome by making use of the trigonométrical transformation.

$$\cos A = 1-2 \sin^2 \frac{A}{2}$$

This is true for all values of A. For the case $A = 10^{\circ}$. the relation becomes

$$10^{\circ} = 1 - 2 \sin^2 5^{\circ} = 1 - 2x (0.087)^2$$

= 1 - 0.01514
= 0.08486

Again, it may be desired to determine the secant of a small angle. At first sight this appears to be deter-minable by taking the reciprocal of the cosine, calculated as above. It will be realised, however, that the degree of accuracy in the value of the secant obtained in this way would be much lower than that of the cosine. In the particular case of 10°, the cosine is 0.98486 (correct to fourth figure), but it is impossible to set the slide rule to anything finer than 0.985 for this purpose and the secant is then 1.015 ± 0.001 . An alternative is to make use of the transformation

Sec. A = $\sqrt{(1 + tan^2 A)}$

$$= I + \frac{tan^{2}A}{2} - \frac{tan^{4}A}{8} +, \text{ etc.}$$

For A = 10°
Sec. 10° = I + $\frac{tan^{2}10^{\circ}}{2} - \frac{tan^{4}}{10^{\circ}} +, \text{ etc.}$

$$= 1 + \frac{2}{2} + \frac{8}{(0.1763)^2} - \frac{(0.1763)^4}{8} + , \text{ etc.}$$

= 1 + 0.01552 - 0.00012 + , etc.

=1.01540

and this may be accepted as correct to the fourth figure cursors are available, such as the Goulding. after the decimal point.

These random examples of application of elementary mathematics to increase the utility of the slide-rule are not by any means exhaustive, but familiarity with them will often be found useful and will also suggest how similar methods may be developed to deal with other forms of slide rule calculation.

USEFUL SLIDE-RULE GAUGE POINTS

		Se	t to
Known Value on Slide	Required Valueon Rule	On	On
strotter / dedo on Source -		Slide	Rule
Square inches	Square centimetres	31	200
Square feet	Square metres	140	13
Square miles	Square kilometres	112	290
	Hectares	$\frac{42}{36}$	17
Cubic inches	Cubic centimetres	106 + 106	590
Square side	Cabic metres	70	3 99
Square, side	Square, diagonal	50	127
Inches, eights	Centimetres	40	127
Feet	Metres	292	89
Yards	Metres	35	32
Miles	Kilometres	87	140
Circle, diameter	Circle, circumference	225	-710
Circle, diameter	Circle, side of inscribed		
	square	-99	70.
Circle, diameter	Circle, side of equal		
Cincle disputer	square	79	70
Circle, diameter	Circle, side of equal equilateral triangle	50	07
Circle, circumference	Circle side of inseriled	72	97
circle, or cumbrelete	square	40	9
Circle, area	Inscribed, square area	300	191
Pounds per square inch-	Atmospheres	1 +0*	33
Pounds per square inch	Water, head, feet	13	30
Pounds per square inch Pounds per square inch	Water, head, metres	33	25
Pounds per square inch	Inches, mercury gauge	25	51
Inches, water gauge		360	13
Inches, water gauge	Inches, mercury gauge	14	1
Inches, mercury	Atmospheres	30	1
Atmospheres	Kilos per square centi-		
	incore	09	92
British thermal unit British thermal unit	Calories	250	63
	Calories per kilogramme	9	5
	Kilogrammetres		5 47
Foot Pounds	Force de cheval	1	73
Pounds per H.P.	Kilos per cheval	000	134
	Kilowatts (B.T.U.)	1 1	100
Watts	Horse-power	5	.0007
Pounds per square foot.	Kilos per square metre		425
Pounds per lineal foot	Kilos per lineal metre	41	61
Pounds per lineal mile	Kilos per kilometre	71	20
Pounds per cubic foot	Kilos per cubic metre	39	625
	Weight in pounds Gallous (imperial)	17	1060
Gallons of water	Weight kilos	17	$ \frac{106}{490} $
Gallons of water Pounds of water (fresh)	Pounds of water (see)	108	490
Feet per second	Metres per minute	1 1	128
Feet per second	Miles per hour	1 00	15
Feet per minute	Miles ner hour	1 044	3
Yards per minute Miles per hour Knots (nouts) per hour	Miles per hour	88	3
Miles per hour	Metres per minute		322
Knots (nouts) per hour	Metres per hour	33	38
Pounds per square inch	Knogrammes per square		1 .
Cubia faat	centimetre		9
	Litres		85
	Cubic metres		39
Gailons			6
Ounces (Avoirdupois)	Grammes		1900
	17 ilow waxa waxa oo	1 650	10
Pounds (Avoirdupois)	Kilogrammes	280	127
Hundredweights	Kilogrammes	5	254
Tons	Tonnes	1 62	63

It is wise to purchase a good quality slide-rule with **cl**ear calibratious. Some of the cheaper models warp badly, making movement of the slide difficult, and accurate reading almost impossible.

For pocket use the circular watch-type of slide-rule is handy and quite accurate. The Halden Calulex and the Fowler circular slide-rules are good examples of this type.

For those whose eyesight does not permit them easily to distinguish the scale markings, magnifying

(To be continued.)

Battery-operated Amplifiers

"Voltage" and "Power" Amplifiers. Single and Two-valve Units. Push-pull and Parallel Outputs Compared. BY THE TECHNICAL STAFF

REAT interest is being shown in circuits for low-power battery-operated amplifiers, and, as blucprints are not available, we are giving the details of those arrangements which we feel will have the widest appeal.



Fig. 1.-A suitable circuit for a power or super-power value.

The stipulation "low-power" is made because we do not recommend to our readers the use of circuits having an output which would bring them within the

medium-power class, when such circuits have to depend on normal dry H.T. batteries for high tension. It is not an economical propositionespecially at the present time -to think of 2 watt and 21 watt outputs. For normal domestic use, and this includes pick-up work, such outputs are really unnecesmany constructors sary; talk about watts in the output stage as if they were free for the asking. Much of this is due to their inability to judge an output of, say, I watt, and their lack of understanding of the power actually required for load reproduction in a normal room.

We do not propose to discuss the claims of the super-

quality enthusiasts; the majority of inquirers seem more concerned with satisfactory reproduction and economical upkeep.

Single Valvers

When it is required to increase the output from a one or two-valve receiver, it is often sufficient to pass the signal through an additional L.F. stage. In fact, if two valves are already in use, it is not advisable to It two valves are already in use, it is not advisable to attempt to add, say, another two stages when utilising battery-operated valves. The reason for this is that the output valve is likely to be overloaded, and the possi-bility of introducing L.F. instability unless very thorough decoupling arrangements are incorporated. The latter is, however, limited in its applications by the normally low H.T. voltage available from batteries.

Assuming the extra stage is wanted to follow on an existing receiver, then a "power" amplifier will be required; this necessitates the use of a super-power valve or an output pentode, as these can handle a reasonably large input—the former better than the

latter—and produce a great increase in power in their anode circuits. On the other hand, if it was desired to amplify, say, the output of a microphone or pick-up before feeding it into an existing amplifier, one would be concerned with securing maximum amplification of the signal, i.e., the input voltage, therefore, a "voltage" amplifier would be required. The difference between these two forms of amplifier is, a power amplifier is solely concerned with delivering a large amount of bower—high wattage, while a voltage amplifier has, as its chief purpose, the *high magnification of its input signal*, without being concerned about power. Such stages make use of valves having a high magnification factor (high pur) which be winter the stages of the stages (high mu), which by virtue of their characteristics cannot accept a large input (grid-swing), neither can they supply any great power from their anode circuit. Power amplifiers, on the other hand, can accept a large input, i.e., that passed on by the preceding voltage amplifiers, but their actual voltage amplification is low. For example, the ratio of the input voltage to the output voltage might be less that 3 : 1. If, however, their power amplification is considered—the ratio of the power supplied to the grid to power output—the figure will be found to be high, as the type of valve used in a power output stage requires a very low power input to produce a large amount of power in its output.



Figs. 1 and 2 show power amplifiers employing superpower and pentode valves, respectively; the latter will give a greater magnification, but, generally speaking, will not handle as large an input as a super-power valve. If, therefore, the receiver pre-ceding this amplifying stage consists of a single valve, it would be advisable to use a pentode to secure the greater magnification. A simple tone corrector is connected across the L.S. terminals in Fig. 2; this is necessary to correct, the slightly rising characteristic, as regards tonal response, of the pentode. The resistance

R can be fixed or variable, the latter allowing the re-sponse to be controlled over a reasonable band width. If a voltage amplifier is required, the circuit shown in



Fig. 3.—A pre-amplifier stage for use where "voltage" amplification is the chief aim.



Fig. 2.—If a higher gain is required, the pentode circuit is recommended. R should be increased to 20,000 ohms if made

variable.

Fig. 3 can be used in conjunction with a valve of the H.L. or H.F. type. A stage such as this is often required to amplify the signals from a pick-up or nicrophone before feeding them into the main amplifier.

It should be noted that Figs. I and 2 make use of an L.F. transformer for the coupling; its primary can be connected directly to the headphone or L.S. terminals of the preceding set, but if the receiver is of the twovalve type, its existing output valve must be withdrawn and replaced by a valve of the H.L. or H.F. type, and the bias adjusted to suit. The output valve should then be plugged into the new output stage. If one wishes to use resistance-

capacity coupling instead of the L.F. transformer, the input circuit of the amplifiers must be modified to that shown in Fig. 4.

Two-valvers

Two valves is the minimum number for an amplifier, when one is considering a unit capable of satisfying the experimental needs of the amateur. A high-gain triode, transformer coupled to a steepslope pentode, will give a very satisfactory overall amplification, quite sufficient for use with microphones of the carbon class and



Fig. 4.—The modification necessary if R.C. coupling is required between set and amplifier.

the majority of pick-ups; and, of course, a H.F. and Det. arrangement or a plain detector stage.

The output obtainable from the majority of batteryoverated pentodes is adequate for any normal room, when an efficient speaker—preferably of the movingcoll type—is used.







PRACTICAL WIRELESS

Fig. 5.—A useful two-valve L.F. amplifier suitable for use with a small receiver, pick-up or microphone.

A suitable circuit is shown in Fig. 5, the valves being a Cossor 210 HL and a Cossor 220 HPT. A volume control is connected across the grid circuit of VI, thus allowing the strength of the input signal to be kept within the satisfactory operating limits of the combination and the output of the amplifier to be controlled. The input circuit, as shown, is satisfactory for direct connection to a pick-up or the secondary of a microphone transformer. If the amplifier is to follow a detector or a H.F. and detector unit, then it will be necessary to use the transformer or resistance-capacity coupling between the set and the amplifier. If one is prepared to sacrifice a little volume, the pentode valve can be replaced by a power or super-power valve, in which case a four-pin valveholder would be used, and the connection from the H.T. line to the auxiliary grid of the pentode ignored. The layout of a two-valve amplifier is so simple that there is hardly any need for us to show a plan drawing. However, here are a few hints for those not familiar with such constructional work. Chassis assemblies are certainly neat, but not essential for this type of circuit. Select a baseboard, say, 7in. by 4in. Fix a valveholder cach and of the base-keeping them on a centre line which should be marked beforchand-leave between the valveholders sufficient space for the L.F. transformer, thus allowing the three components to be fixed nearly in line. This will ensure the wiring being short and direct and the anode and grid circuits kept well apart. A small panel can be fixed to one of the longest sides

A small panel can be fixed to one of the longest sides of the baseboard, and used to mount the input and output terminals, the volume control and the variable

tone control, if a variable resistance is used for that circuit. As mentioned before, the resistance can be fixed if the other type of component is not to hand. When the amplifier has been assembled and tested, it would not be a difficult matter to make a small case to house it.

Push-pull and Parallel Outputs

If the components and valves are available, and amplifier having a push-pull output stage forms an ideal arrangement, provided ene is not too concerned about H.T. current consumption, as might be the case when a good H.T. eliminator is in use.

A circuit is shown in Fig. 6, suitable valves being H.L., H.F. or L.F., and two super-powers. The two voltage amplifiers are advisable to ensure full loading of the output valves under all input conditions, the volume control being provided to prevent overloading and, of course, normal control of the signal from the point of view of volume.



Fig. 7.—Showing the connections necessary for coupling two valves in parallel.

Resistance-capacity coupling is used between. Vr and V2, but the output from the latter is fed to the output valves by means of a suitable push-pull input trans-former. This component has a centre-tapped, or, as in some models, a split secondary winding, and a low ratio

between secondary and primary. If such a transformer is not obtainable, use can be made of an ordinary L.F: transformer having a ratio of; say; 3 : 1, by connecting across its secondary winding two 0.25 megoha resistors—connected in series. The junction point formed between these two components is then used in the same manner as the centre-tapping on the correct type of transformer.

The output obtainable from a push-pull stage is, approximately, just over twice that of one valve of the same type; the same result as regards power output can be achieved by connecting valves in parallel, Fig. 7, and it is interesting to note that, assuming the same operating conditions to exist in each case, the parallel arrangement would require only half the input of those in push-pull or, the other way round, valves in push-pull will give the same output as two valves in parallel, provided twice as much input is applied.

Before it is thought that this comparison shows a cheap way of obtaining a large output by using valves in parallel, the additional details should be noted. When using triodes in push-pull, second harmonic distortion is eliminated, and as this is a troublesome factor in L.F. amplifiers, the advantage offered by the push-pull system is, therefore, greatly enhanced. In practice, it means that valves in a well-designed push-pull output stage can be made to deliver much more power before distortion creeps in than an output using a single valve or wo valves in parallel; therefore, to come back to our original, comparison, it must be modified to read: "Valves in push-pull have a greater undistorted output than valves in parallel.'



URING the past month correspondence from mémbers has not been so heavy as previously. This does not mean that interest is lagging; as we see it-guided by our own experiences, and remarks in those letters we have received—it indicates, that service in the Forces, war work and Civil Defence duties leave little time these days for correspondence. In spite of these arduous essential duties, members still find the time to keep their hand in, their interest alive and in step with such progress as is possible. This is a good sign. It shows that, however hard are our tasks, war restrictions and conditions, we are still able to retain-without sacrificing one atom of our war effortsthe right sense of proportion and the true amateur spirit.

If your letters to us have to be fewer, make up the deficiency by making them as interesting-to all members -as possible.

Swaythling, Member No. 6,410

I FEEL that it is time I wrote to you relating my activities in the short-wave sphere. My den is housed in my bedroom, so I have to make the utmost use of the little available space. To this end I use a three-tier dinner wagon for my table, etc. On the top that is a ranged a microphone, mixer of my own design, battery-operated Class B one-valve short-wave receiver (which has provision for H.F. unit later) and a morse oscillator. The second shelf carries battery charger, cabinet containing stand-by s.w. Rx and some tools. On the third shelf there is a large box of spares and many tins of nuts, bolts and washers, etc.

"To the left of the dinner wagon stands a homemade bookcase containing PRACTICAL WIRELESS, and many other books. On top of the case is a large cabinet of spares, and also a two-valve all-wave receiver and a two-valve medium-wave set. All apparatus, except the battery charger, is home-constructed.

My friend and I have spent several all-night listening. periods together, but without good results. From experience gained I find the best listening period to be from 16:00 to 01.30 hours. I am not in a good locality, being in a valley, but have tried to overcome this with the following aerial arrays, all of which are running, approximately, west to east: (r) 66ft. C.T. 33ft. lead in; (2) doublet r6ft. 6in. sections and, lastly (3) 33ff. tapped 8(t. 3in, from west end. All these are connected to a three-way switch which simplifies acrial changing. these arrays I have received the following: WNBI, these arrays I have received the following: WNBI, WRCA, WGLA, WGEO, WCBX, WRUL, WLWO, WCW, FET22, HVJ, OIE, OFD, SBO, RWG6, RNE, RAL, JZI, JZI, MTCY, XGOY, TAQ, TAP, VLQ, VLQ2, FZR, FZI, CSW6, VUD4, CR7BE. Best wishes to all concerned in keeping PRACTICAL WIRELESS going."

[Your drawing, showing the lay-out of your gear, arrived too late for inclusion with your letter. However, it may be possible to reproduce it in the next issue. You have certainly made the most of the small amount of space available .- HON. SEC.].

Suffolk, from Member No. 8,095

HERE are a few details of my s.w. receiver, which I have just completed. Like many other members, I now have very little time for experimenting, but having been without a short-wave set for several months I decided to make up a simple set, which I am pleased to say gives quite good results.

The set is constructed on an old Cossor 234 Melody Maker chassis, utilising some of the original components. Apart from substituting a pentode for the output triode, the L.F. section was not changed. The detector is resistance-coupled to the first L.F. valve, as shown in the circuit (Fig. r). Both of the original tuning con-densers (.0005, mfd.) are used for tuning and reaction. As previous experiments with commercial short-wave (Continued on page 36.)



The "Fluxite Quins" at work.

" I shall have to climb down!" OI yelled out.

" Sixty feet to the ground—just about," Said EH in his car.

"Here's the FLUXITE—don't fear. "Just stay where you are—you young cout."

See that FLUXITE is always by you—in the house—garage workshop — wherever speedy soldering is needed. Used for 30 years in government works and by leading engineers and manufacturers. Of Ironmongers —in tins, 4d., 8d., 1/4 and 2/8.

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ON SMALL PARTS...

N countless instances quite intricate pieces of apparatus are wholly dependent on the proved reputation and reliability of their - component parts.

All products from the House of Bulgin are preeminent for superior design and workmanship, and every article bearing our Trade Mark has to pass exacting and exhaustive tests during the course of its production.

We ask the kind indulgence of the public in delivery until peaceful conditions return.

ALWAYS DEPEND ON



coils have not proved very satisfactory, I decided to make one of my own. A suitable coil former was not to hand, so I utilised a piece of ebonite fin. by 2 in. by in (Fig. 2). The grid winding consists of 13 turns of .028 tinned copper wire (22 s.w.g.), the turns being spaced in apart. The reaction is wound between the turns of the grid winding, and consists of 13 turns of (26 s.w.g. ?) enamelled wire taken from the secondary winding of an old output transformer. The leads connecting the coil are short and thick, and also act as supports. A



Fig. 1.-The three-valuer described by Member 8,095. a home-made coil and H.F. choke. It utilises

crocodile clip, on a short earth lead, is used to clip on to the coil for various wavebands. Although crude, the coil is reasonably selective, and covers a range of about

"The H.F. choke is based on suggestions made in an old issue of PRACTICAL WIRELESS, and is wound in five sections of 30 turns each on a cardboard case of an old No. 8 torch battery. "The American stations come in well from early

"The American stations come in wen from early afternoon to late evening. Among the various stations received are WLWO, Cincinnati, Ohio, 25.6 m.; WRUL, Boston, 25 m.; WGO, New York City, 20 m.; Radio Centre, Kuibishev, U.S.S.R., 20 m.; Radio Centre, Moscow, U.S.S.R., 41.1 m.; WGEA, Schenectady, 19 m. I heard WGO for the first time the other day, giving a news bulletin in English, at about 2 p.m. B.S.T. WCBA was also heard round about the same time. This station was also heard round about the same time. This station

"Here are the latest times, etc., of English transmis-sions from Moscow, which came into operation on October 1st: 4 p.m., 25.36 m.; 5 p.m., 42.98 m. 6 p.m., 8 p.m., 9 p.m., 10 p.m. (all B.S.T.), 41.1 m.; 11 p.m. B.S.T., 39.68 m., 40.76 m., 41.1 m., 42.98 m. 50.93 m."

Short-wave Receivers

T is a well-known fact that much greater care must be exercised in the construction of short-wave receivers than in that of sets designed for reception of the normal broadcast bands. This is particularly true with respect to the various, earth connections. While experimenting with a S.W. receiver it was found that the circuit proved very unstable unless a lead was connected direct from the terminal attached to the moving vanes of the tuning condenser to the earth terminal of the coil. When the set was first assembled this terminal was unused, it being assumed that a sufficiently good contact between the coil and con-denser, was effected through the metal bracket, attached proved quite satisfactory for medium and long-wave reception, but immediately the receiver was switched over to the short waveband marked instability was experienced. It is therefore emphasised that members

should strictly adhere to the designer's layout and wiring when constructing a short-wave receiver. It is also pointed, out that an efficient earth connection is very desirable, when listening on the short wavebands. A bad earth connection causes hand-capacity effects and instability.

Electric Soldering Irons WHEN receiver construction is being carried out and an electric soldering iron is used, some constructors

place the iron on a sheet of metal or other metal support during the wiring process. It may be found that the metal will, in such a case, conduct away a large portion of the heat of the iron with the result that when a connection has to be made the iron is not sufficiently hot and has therefore to be held in contact with the component being soldered for an undue period, with possible damage to that component. A better plan is to rest the iron so that the tip and a portion of the iron which carries the heating element, is surrounded by air, but if the iron is not wanted for fairly long periods of time, a series resistance should be included to avoid undue waste of current and possible damage to the iron.

Wiring Supports

IN some receivers a long bare wire is sometimes called for to act as

a common connecting point—bus-bar is the name generally applied to this. If this tinned copper wire is being used there are two schemes which will give rigidity and avoid risk of damage due to a sagging wire. In the first case-small stand-off, insulators may be attached to the chassis and the wire stretched between them. An alternative scheme is to use two strands of the thin wire and twist these by holding in the jaws of a twist drill. Stretch them slightly while twisting, and they will provide a rigid connecting link which can run for quite a considerable distance without risk of damage.



MANY members may wonder why their circuits have not been reproduced on these pages. To avoid any misunderstanding, we wish to stress the following. The bare theoretical circuit of a receiver is not of general interest. All component values must be given, and, if any particular section of the set is of special importance, full details are essential. If home-made coils are used their constructional data must be provided. We will welcome details of members' sets; provided that full information is given.

Impressions on the Wax

Review of the Latest Gramophone Records

H.M.V.

 $\mathbf{T}_{\mathbf{N}}^{\mathbf{WO}}$ rzin. records in the *DB* series open up my list of selected recordings for this month. The first is DB2846, which gives the two latest of the many famous recordings by that great artist of the violin Heifetz. recordings by that great artist of the violin Hentetz. For these outstanding performances he selected "Chant De Roxane" from the Opera "Le Roi Roger" (Szy-manowski trans. Kochanski) and "Dance Espagnole" from "La Vida Breve" (de Falla trans. Kreisler). He is accompanied on the piano by Emanuel Bey. The artistry and technique of Heifetz are superb, and the two compositions he renders gives him anple scope to reveal in detail these outstanding onalities. HMV reveal in detail these outstanding qualities. H.M.V. DB2846,

The other record, H.M.V. DB6135, is a fine recording by the San Francisco Symphony Orchestra—conducted by Pierre Monteux—playing "Pièce Héroïque "—Part I and Conclusion. The Ioin. records included in my selection form what might be classified as a light "wide selection what many be classified as a near wide appeal" programme. For example, I open up with a selection which should meet with general approval. I am referring to *H.M.V. B9294*, on which will be found Parts I and 2 of "Wild Rose" Selection—played by the New Mayfair Orchestra under the direction of Debray Second I for direction and the main direction of Debroy Somers. I found it pleasant entertainment, easy to listen to and enjoy. It introduces Overture; You Can't Keep a Good Girl Down; Look For the Silver Lining; On With the Dance; Whip-poorwill; Sally; The Church Round the Corner, and Finale. Remember the number, H.M.V. B9294.

For a vocal with piano, I include Joyce Grenfell singing "I'm Going to See You To-day" and "There is Nothing New to Tell You," with Richard Addinsell at the piano. H.M.V. B9295. For the dance enthusiasts; I would suggest The R.A.O.C. Blue Rockets Dance Orchestra playing "Desert Patrol"—a quick-step, and "The Great Little Army," a fox-trot march. These are two rood numbers—the last being a chythmic accomment Great Little Army," a fox-trot march. These are two good numbers—the last being a rhythmic arrangement by Tommy Keith—played in that fine manner for which the Blue Rockets are becoming so well known. $H.M.V. DB_{5775}$. A waltz is on the next record for the turntable, and it has been recorded by Joe Loss and his Orchestra, the piece being "Where the Waters Are Blue," and this is followed by a slow fox-trot entitled "Where In the World," both being on $H.M.V. BD_{5773}$. To finish off this little H.M.V. programme I have picked "Red Lips"—rumba fox—and "London Conga," two distinctly attractive pieces played by Don Marino Barreto and his Cuban Orchestra and recorded

Marino Barreto and his Cuban Orchestra and recorded on H.M.V. BD5776.

Columbia

TWO superb artists have collaborated for their first solo recording, and it is given on *Columbia DX* 1094. They are Henry Holst—violin—and Gerald Moore— piano—therefore there is little call for me to make any comment, as their performance is all that one would expect from such talented artists. Holst selected for his Columbia début—as a soloist—that lovely piece by Delius, "Legende," and he plays it with masterly under-standing of Delius's magic. Gerald Moore, the planist in this recording, forms the perfect collaborator, and his performance reflects his birth credition Columbia DX rog4: "Delius-Legende in E Flat for Violin and Pianoforte," Two Parts. Schubert's "Impromptu in A Flat" forms the subject

of Louis Kentner's latest record. The A Flat Impromptu is delightfully Schubert. It takes one through the whole is denginerary schuber of the tanks on a subschuber of the tanks of tank the great Schubert. The performance of Louis Kentner leaves nothing to be desired. Columbia DX1093. Two Parts. In the *DB* series I recommend the following : *DB*2093, on which that lovely soprano Isobel Baillie has recorded—with Gerald Moore at the piano—"Stript of Their Green" and "I Saw That You Were Grown So M_{12} .

High." From "Wild Rose," Jessie Matthews has selected "Whip-poor-will" and "Look For the Silver Lining" for her two numbers on Columbia DB2094, and she is as delightful as ever. Debroy Somers conducts the orchestra which accompanies her.

orchestra which accompanies her. Another vocal is FB2858, On this Turner Layton renders that Irving Berlin number "Be Careful, It's My Heart," which is featured in the film "Holiday Inn." On the other side of the disc, Turner offers us "Three Minutes of Heaven." Two good numbers in true Turner Layton style. Columbia FB2858. Now for a couple of dance records. The first is a quickstep "I've Seen You Before." and "Where in the World "—slow foxtrot—both played by Victor Silvester and his Ballroom Orchestra. on Columbia FB2856

and his Ballroom Orchestra, on Columbia FB2856. Personally, I can always listen to Victor's music ; maybe this is because super rhythm does not appeal to me as much as the strict dance tempo. However, it would not do for all of us to have the same tastes, but I am pretty sure that Victor has a very large following. Carroll Gibbons is another who always finds a place

near the front of my dance record library. I once referred to his touch as "lazy fingers"—well, most of his presentations are remote from the more hectic renderings which are so often classified as dance music. The two pieces he has selected for recording with the Savoy Hotel Orpheans are "It Isn't a Dream Any More" and "Conchita Marguita Lopez" a fine fox-trot. The last number, by the way, is featured in the film "Priorities of 1943." Columbia FB2853.

Parlophone

A^T the top of my Parlophone list is Richard Tauber on A Parlophone RO20515, singing two very well known songs in English, namely, "Good-bye" (White-Melville and Tosti) and "My Dreams" (Weatherly—Tosti). Orchestra—under the baton of Henry Geehl—accompaniment.

paniment. For the dance fans, there are three records, two of them being by Geraldo and his Orchestra. On F_{1943} he offers a fox-trot "Jingle Jangle Jingle" which is featured in the film "Forest Rangers," and "Hey Mabel" also a fox-trot. On his other record, *Parlophone* F_{1944} , we hear him playing "Ganzonette" and "Sweet Eloise," two more fine fox-trot numbers. Nat Temple, by the way gives a very fine claringt solo performance. by the way, gives a very fine clarinet solo performance in "Ganzonette."

The super rhythm enthusiasts will be pleased to hear that Harry Parry and his Radio Rhythm Club Sextet, have two numbers on *Parlophone* R2851; they are "Rock It Out." and "I Can't Dance," the latter having vocal refrain,

Decca

D^{VORAK'S} Humoreske, Op. 101 No. 7, arranged Kreisler, is a delightful piece, and Ida Haendel with Adela Kotowska at the piano-renders it with such skill and feeling that the recording is superb. It will be found on *Decca* M521, together with "Caprice Viennois" by Kreisler, also played by the same artist. This is a record of note and I recommend it to all who enjoy violin solos of the first order.

Joan Taylor—soprano—gives a delightful rendering— on *Decca M*523—of two ever popular songs. They are "My Hero" from "The Chocolate Soldier" and "Vilia" from "The Merry Widow." On *Decca F8*203, we have another "Humoreske," but this time by Tchaikovsky (Op. 10 No. 2) and played by (Continued on page 40.)

Musical Frequencies

The Relationship Between Mathematics and Music. By CHARLES WILLIAMS

SINCE the mathematical study of frequencies plays such a large part in wireless theory, it is of considerable interest that there is a mathematical

basis for musical frequencies also. Musical intervals, were not, of course, invented by the aid of mathematics, but once a scale of musical intervals has been developed it is legitimate to inquire whether there may not be a mathematical reason why these particular intervals have been found pleasing to the ear, and a suitable basis for the construction of harmonies.

The ancient Greeks had three systems of intervals, based on groups of four notes called tetrachords. In one of these systems, the diatonic, the intervals downward from the upper note of the four, were in succession, a tone; a tone, and a senutone; or counting upward, as we are accustomed to do, semitone, tone, tone. It is from this system that our present diatonic scale has developed.

It is said that Pythagoras, as early as the sixth century B.c., discovered that there was a peculiarly simple mathematical relationship connecting this group of notes.

Frequency Ratios

The fundamental mathematical basis of harmony is In order that two notes may be in harmony there this. must be a simple ratio between their frequencies. The simpler the ratio, the more perfect the concord. On this basis it is possible to give a mathematical account

of the major diatonic scale. The simplest possible ratio between two frequencies is 2 : 1. This is the ratio of the octave. The concord is here so perfect that the two notes are recognised as in a sense identical and are given the same name. Thus on the piano we have a succession of notes named C, each having twice the frequency of the one below. The next simplest ratios in order are:

3:2 4:4 5:3 5:4

which, C being the keynote, gives the notes G, F, A, E. Taking, in order to avoid fractions, 24 as the frequency of our keynote, and calling this C, we get the series :

The successive frequency ratios of the four notes E, F, G, A, are: 32:30, 36:32, 40:36, or more simply 16:15, 9:8, 10:9. Here the ratios 9:8 and 10:9 are nearly equal, being equal to 1.12... and therefore regard the intervals F to G and G to A as being approximately equal and each equal to 1xi2... and therefore regard the intervals F to G and G to A as being approximately equal and each equal to twice the interval. mately equal, and each equal to twice the interval from F to G.

Diatonic Scale of Tones

If we call the interval from F to G a semitone, we have in this group of four notes E, F, G, A, the diatonic tetrachord with intervals, semitone, tone, tone. Its advantage harmonically is that each of the four notes has one of the four simplest possible frequency ratios with the keynote.

We have now to consider the intervals C to E, and A to C. The ratio of the frequencies of E and C is 30 : 24, which is the same as 5 : 4. We note that

$$\frac{5}{4} = \frac{10}{9} \times \frac{9}{8}$$

which suggests that we can put a note D between C and E such that its interval from either will be a tone, i.e., 9: 8 or 10:9. This gives:

$$C D E F G A - C$$

24 27 30 32 36 40 - 48.

So far there has been no alternative to the suggested

development if the simplest ratios were to be preserved, but the interval A to C, which is equal to three semitones, might be treated in more than one way. In fact, in the major diatonic scale it has been divided into a tone and a semitone. Thus :---

$$\frac{48}{40} = \frac{6}{5} = \frac{9}{8} \times \frac{16}{15}$$

which gives 45 for B.

The complete octave is therefore :---

Note			C.	D	E	F	G	A	B	e
Relative	frequer	ıcÿ	24	27	30	32	36	40	45	48
Ratio	a -565	and the	9(00	ro	1	6 9		o .	9.0	16
Interval		•••	Т	T	S	5]	r i	ŕ	Т	S

where T stands for a tone and S for a semitone.

What has been established so far is that the diatonic scale of tones and semitones in the above order appears to be, on mathematical grounds, the most satisfactory scale on which to build harmonies. If music in one key only were required, all that would be If necessary, in the construction of a keyboard instrument such as a piano, would be to repeat this series of intervals as often as desired to cover the compass of the instru-ment. This would give a keyboard with the white notes only, and all the intervals could be made to have exactly the frequency ratios given above.

Intermediate Notes

If, however, another note than C is to be taken as a keynote, and the series-tone, tone, semitone, etc., is to start from that note, additional intermediate notes will be required. The only way to make a practical keyboard fulfilling this requirement is to treat all tones as exactly equal intervals, and each of them as equal to two semitone steps.

A keyboard can then be constructed on which every semitone is represented. Five additional notes will be required in each octave, to divide the five tone intervals into semitones. The intervals E to F and B to C are already semitones and do not need division.

The octave will then consist of twelve semitone in-tervals, not all the notes being required, however, in any one scale. The five additional notes are, of course, the black notes of the plano. This procedure is fortunately possible because the ear does not detect slight departures from a simple frequency ratio. Two notes whose fre-quency ratio is very nearly 3:2 form practically as satisfactory a concord as if the ratio were exact.

If all the semitones are to be equal, what frequency ratio must correspond to a semitone? An interval of an octave, which is twelve semitones, doubles the frequency. Therefore, each of the twelve steps of the geometrical frequency progression must be equal to the twelfth root of two which is, calculated to five figures, 1.0595. Note how near this is to 1.06, which we previously found for the semitone interval.

Geometrical Progression

The frequencies of the notes on the piano, when both black and white are included, therefore form a geo-metrical progression whose common ratio is the twelfth root of two. A tone consists of two steps in this pro-gression, and a semi-tone of one step.

It remains only to consider how far this tempered scale differs from the theoretical and true diatonic scale. The degree of difference is best exhibited by a table in which the keynote is represented by frequency one,

one negue	to it represented	by more unity 0.	1
Note	Tempered	Diatonic	
	Frequency	Frequency	
C	I.	I.	
be D	1.1225	1.1250	
-		1	

E	1.2599	1.2500	
\mathbf{F}	1.3348	I.3333	
G	. I.4983	1.5000	
Α	1.6818	I.6666	
B	1.8877	1.8750	
C	2.	2.	

Even the best piano tuned is therefore slightly out of tune in relation to the natural diatonic scale. It is a remarkable fact, and a fortunate one for musical instrument makers, that the powers of the twelfth root of two approximate so closely to the simple ratios of the diatonic scale. If they did not the construction of a piano keyboard would be impossible. In the case of the interval C to G for example, the error is hardly greater than one in ten thousand, which is quite appreciable to the ear.

All the frequencies here given are purely relative. The pitch of middle C is chosen arbitrarily, and whether 256 or 261 or any other number is taken the frequencies of the other notes can be worked out from the ratios already given.

SCUSSION

The Editor does not necessarily agree with the opinions expressed by his correspondents. All letters must be accompanied by the name and address of the sender (not necessarily for publication).

Tenth Anniversary of "P.W." : Crystal Reception SIR,-I hope you will get me right when I say I see **D** nothing very wonderful in the survival of PRACTICAL WIRELESS and its Tenth Anniversary.

Its merits have been so outstanding, as compared with its competitors, that I don't see how it could fail to survive and flourish. It is packed from cover to cover with the most valuable technical information and assistance, and your own "Refresher Course in Mathe-matics" alone makes it worth far more than its published price.

This is a subject which, I must admit, has been a bit beyond me, but I know enough to realise what is being put into these articles, and I have little doubt that some of the Great Intellects of the Brains Trust may be

some of the Great Intellects of the Brains Trust may be carried out of their depth in this direction. "Sutor ne supra crepidam judicaret." Let them-stick to what they are more familiar with, such as "Flies on Ceilings," etc. By the way, I was interested to note F. G. Rayer's comments on "Loud-speaker and Crystal Set" in Open to Discussion Page, but I think he has missed the point. We knew many years ago that bhones blaced in a dish would ago that phones placed in a dish would slightly increase volume from a crystal set and, as he says, a sensitive loud-speaker of ordinary type would also give "bedside" reception, but not sufficient to fill a small sitting-room at comfortable strength.

A speaker constructed on the lines I suggested would, I think, overcome this defect, but only actual experiment could settle this question, and I had not the facilities. Some day the idea may be tried out.—K. T. HARDMAN (Birkenhead).

Simple Morse Oscillator

SIR,-Having a little time to spare recently I decided to experiment with a morse oscillator I had built some with a mark occurator I had only some time ago from a simple circuit, shown in the accompanying diagram. The chief components used were a Varley "Nicklet" 3.5/I L.F. Trans. (T1) and a Midget R.I "Parrafeed" L.F. Trans. (T2). The latter was

K. FAITHER L.F. Irans. (12): The fatter was included with intention of increasing volume, also to obtain a higher frequency. C.I is a .00075 mfd. "Polar" solid dielectric condenser (any other value could be used) for the frequency control. Keying is obtained by short circuiting the primary of the trans-former L. with a hear cated part here. former Tr with a back contact morse key. The valve can be a S.G. or an L.F. output pentode.

The anode voltage is lower than the screen voltage,

depending on the valve in use, and is best decided by experiment .- P. KING (Swindon).

Station HCJB

SIR,-I have just received a communication from Station HCJB, "The Voice of the Andes," in Quito, cuador, and, in view of the fairly widespread popularity of this station among DX enthusiasts, the following details may be of interest.

A complete history of the station since its inception in 1931 is given and relates how in that year, with a staff of three, the first "Radiodifusoras HCJB" was installed and inaugurated. The description traces its rise from a modest 200 watt local transmitter to that day in 1940 when, after two years of planning and construction, the new 10 kW station was opened by the President of Ecuador and the signals from "La Vaz de Los Andes" were at last destined to reach every corner of the globe.



Circuit diagram of a simple morse oscillator, described by P. King, of Swindon.

The new verification card of the station is included in the communication. It portrays the Ecuadorian colours—red, blue and yellow—and displays a photo-graph of the station and its beam antenna, behind which lie many of the Andean peaks. There are actually six stations using the same call. These are:

- 1. The international on 12.4 mc/s-10 kW with a four-element rotary beam.
- 2. A provincial on 4,107 kc/s—1 kW. 3. A local on 970 kc/s—1 kW with a three-element beam

- An amateur (HC1JB) on 28,370 kc/s.
 An experiment on 27,985 kc/s.
 A mobile on 20 metres with a half-wave vertical antenna.

The 12.4 m/c station is on the air every day except Mondays and broadcasts programmes in five languages. Its exact locality is to the north of Quito. It holds a unique position in being the first station constructed chiefly for the purpose of religious devotion, and on account of this bears the title of the "Pioneer Missionary Broadcaster."—JOHN W. MACVEY (Kelso).

S.W. Listening

SIR,—The comments and opinions expressed in the November issue of P.W. by F. G. Rayer are most interesting. I assume he questions my statement in the general sense. The facts, however, are as stated. With this and several other instances in mind, including correspondence devoted to the same subject in a pre-war contemporary in which the solutions were a special coil, or alternatively removing one complete turn from grid winding, a friend of mine who uses the same value of tuning capacity as F. G. R. experienced the same trouble, and had to wind a special coil on a standard former. In all instances, including mine, the receivers have been beyond question in every way, including the maximum capacity of tuning condenser. Band spread is used generally, but this condenser was cut out during tests. I do not question your correspondent's results, and the conclusion reached was that whilst the rated approximate minumum was 9 metres in practical use, the actual minimum was around 12 metres. The question of dead spots does not arise, as my receivers are entirely free from these effects, and smooth oscillation is obtainable over the full tuning range-below 10 metres to 180 metres. Entire freedom from body capacity effects is also usual. In view of his remarks relative to aperiodic windings, has your correspondent removed interwound aperiodic windings, and rewound them to suit himself? Also, does he use a standard low loss commercial coil base ? Whilst I agree in some measure with his comments I think that taking into account the efficiency of modern six-pin coils used, and inductively coupling to H.F. circuits, and as a form of doublet coupling it seems that they are designed as a compromise. The aperiodic coupling being much too tight to allow oscillation of o-V-I type detector circuits. I am fully aware as to how simple it is to obtain satisfactory and true aperiodic coupling with home-made six-pin coils. These coils are designed for the job, and not as a compromise. I do not alter commer-cial coils, or favour this practice. I prefer to follow the makers' instructions, and judge them on the results obtained. With a home-made coil I find it possible to cover the 10 metres band on a standard S.W. receiver, and the same applies to 5 metres. A suitable H.F. choke, of course, is incorporated. The opinions of other readers are invited. A study of the correspondence so far rather indicates that opinions will be divided. The question is can you cover the to metres band using a ocor mfd. tuning condenser, and a commercial coil rated at approximately 9 metres minimum range, including and excluding band-spreading condenser? The examples quoted by me are not confined to one make of coil, but three different makes, and with losses reduced to the minimum in detector circuits the obtainable minimum in all cases is above to metres. Valves used are standard HR210 triodes. As modern sponsored designs always include a series aerial condenser of semi-variable type, it seems that designers generally come up against this snag.—A. W. MANN (Middlesbrough).



IMPRESSIONS ON THE WAX

(Continued from page 37)

that very fine orchestra-conducted by Montague Birch - the Bournemouth Municipal Orchestra. On the other side of the record, they give Schubert's "Marche Militaire" and, needless to say, their performance is outstanding in all respects.

The Royal Air Force Dance Orchestra have selected for their recording this month "Conchita Marquita Lopez" and "Be Careful It's My Heart," two numbers which are well to the fore and in this instance, well played. Decca F8194.

Charlie Kunz and his Ballroom Orchestra, with, of course, himself at the piano, offers—in strict tempo and without vocal chorus—two fox-trots "Look For the Silver Lining" and "I've Scen You Before," on Decca F8199. Two good tunes nicely presented.

A record which is fine for our own entertainment (when we require something light and tuneful), and ideal for the many thousands of war workers is *Decca F8209*. It is entitled. "Music While You Work " Series No. 1 and 2. The first side of the record, that is No. 1, records a good selection from "The Merry Widow" and introduces "Vilia," "The Merry Widow Waltz" and "The Merry Widow March." On the other side of the disc (No. 2), another bogular selection is provided this time under the "another popular selection is provided, this time under the title of "The Maid of the Mountains." This introduces "Paradise for Two," "A Bachelor Gay Am I" and "Love Will Find a Way." The Victory Band make the macrification of the anisother partormance recordings a thoroughly enjoyable performance.

To close my Decca list, I have selected Ambrose and his Orchestra playing "Jingle Jangle Jingle" and "Rolleo Rolling Along" a fox-trot and waltz respectively. I recommend this to all dance enthusiasts; it is Decca F8212.

Rex

"SANDY Joins the Home Guard" is a merry little-comedy sketch—in two parts—from that popular comedian Sandy Powell. In it, he introduces the song "The Home Guard," and puts over the comedy stuff in his own inimitable manner. Rex 10154.

On Rex ro150, Issy Bonn-with orchestral accompani-ment-records "Where in the World" and "This is. Worth Fighting For."

Finishing once again on dance tunes, Jay Wilbur and his Band, on *Rez* 10151, plays "You Walk By" a fox-trot, and "Farewell to Dreams"—waltz.

PRIZE PROBLEMS

Problem No. 438

THE two-value battery receiver which James was using had suddenly developed a peculiar acratching sound at one point on the training scale. He thought that this was, perhaps, due to condenser vanes short-oricuits, and accordingly proceeded to lest the receiver to that end. He connected a voltmeter in series with a 1.5-volt cell and joined this across the two terminals of his variable condenser. He found that a reading was obtained on the meter, no matter where the condenser was adjusted, and he therefore assumed that the condenser was shorting. What was wrong with his test?

Three books will be awarded for the first three correct solutions opened. Mark your enrelopes Froblem No. 433 and address them to The iEditor, Phacritat Minntass, George Newnes, Ltd., Tower House, Southauption Strett, Strand, London, W.C.Z. Entries must be received by the first pret Monday, November. 1916, 1942.

Solution to Problem No. 437

Jones overlooked the fact that the anode resistor of the B.C. stays acted to some extend as an H.F. chuke, and also restricted, the applied voltage on the anode of the delector. When the L.F. transformer was substituted, no H.F. choke effect was obtained, and a much higher voltage reached the anode, these two items combining to produce instability.

The three following readers successfully solved Problem No. 436 and books have accordinally been forwarded to them. funr. F. H. Watts; R. Barlow, Ulverston : Rugh Wilker, Montrose.

Replies to ueries

RULES

We wish to draw the reader's attention to the fact that the Queries Service is intended only for the solution of problems or difficulties arising from-the construction of receivers described in our pages, from articles appearing in our pages, or on general wireless matters. We regret that we can the construction of receivers described in our pages, from articles applied to obvious reasons: —
 (1) Supply circuit diagrams of complete multi-valve receivers.
 (2) Suggest alterations or modifications of receivers described in (3) Suggest alterations or modifications to commercial receivers.
(4) Answer queries over the telephone.
(5) Grant interviews to querists.

aur contemporaties. (3) Grant interviews to querists. A stamped, adviressed envelope must be enclosed for the reply. All sketches and drawings which are sent to us should bear the name and address

of the sender. Requests for Blueprints must not be enclosed with queries as they are dealt with by a separate department. Send your queries to the Editor, FRACTICAL WIRELESS, George Newnes, Ltd., Tower House, Southampton Street, Strand, London, W.C.2. The Coupon on page ill of to rear must be enclosed with every query.

Heater Leads

of the sender.

"I am making up an A.C. receiver and am rather anxious regarding the wire to be used for wiring up the heaters. In some receivers which I have looked at I find that solid wire is used, while in others flex has been employed. Does it matter what type of wire is used, provided that the current-carrying capacity is ample for the total heater circuit ?"—T. R. (Kenton.)

Is ample for the total nearer circuit : -1. R. (Remot.) THE usual reason for using flex is that it is more easily bent to follow the run of wiring. If solid wire is employed for a given current rating, it will be heavier than the stranded or flexible wire, and therefore more awkward to twist and bend. Furthermore, the flex must be bared at the points where it makes contact with the various valveholders in the wiring and this means that there are two junctions at each valve heater pin. This may lead to difficulties as there will be an undue thickness of wire at that point. By using solid wire of suitable diameter it may be soldered to a valve pin, insulated sleeving slipped over ready cut to a snitable length, and thus only one joint has to be made and this is generally simpler for the amateur. be made and this is generally simpler for the amateur.

Anto-transformer

"Can you tell me what an auto-transformer is ? I thought these things were only used in A.C. mains motor-generators, but an old circuit which I have turned up for radio specifies an auto-transformer for L.F. coupling, and I cannot see how this would work."—F. N. (Lewisham.)

THE auto-transformer consists-merely of a tapped choke, used THE anto-transformer consists merely of a tapped choke, used for L.F. coupling and fed in the usual way from the anode of the preceding valve through a condenser to the tapping. One end of the choke is joined to the grid and the other to G.B. (or earth)? By changing round the connections to the ends two different ratios are obtainable. The portion of the winding between the coupling condenser and earth acts as the primary, and the the large are accounted and thus the relief a darge and the total winding as secondary, and thus the ratio is depen-dent upon the amount of wire between coupling condenser and earth

Potentiometer Connections

"I am including a potentiometer in a test circuit, but only require to use the unit as a variable resistance. One side of the element is joined to earth, but I am uncertain whether I should join the arm of the control also to the other end of the element or whether this latter point should be left unconnected."—B. U. (Teddington.)

IN most circuits it is unnecessary to connect one side of the element to the arm when using a potentiometer as a variable resistance. "The "dead end" will be of no consequence, and the only effect of connecting both together is to short-circuit the unused portion of the element. In some special circuits it might be desirable to do this, especially if a wire-wound component is employed but for all expressions to the two connections along employed, but for all general purposes the two connections alone will suffice.

H.F. or L.F. ?

"At present I am using a single-valve (detector) receiver, but naturally I get a limited field of reception. I am prepared to go to the expense of adding one further stage of amplification, but I have heen told that if I add L.F. I shall not get any more stations, whereas if I add H.F. I shall get more but not any louder. I should be glad to know exactly what will happen with either form of amplification and which is best for my purpose."—W. H. (Oxford.)

IN theory the addition of an L. F. stage will merely increase the volume of signals received by the detector, while the addition of H.F. will increase the signals present on the aerial and thus give you a greater field of reception. In practice, however, there are differences in this. With your detector you will probably ind that signals are present which are extremely weak

and probably can only be discerned when you concentrate seriously—and in some cases they may be just inaudible. The addition of an efficient L.F. stage will bring these up to audibility and thus you will probably hear stations at present not received if you add L.F., in addition to the obvious strengthening of those signals which are now heard. The fitting of H.F. will, besides bringing in stations not now received, often so increase the strength of those extiner you do hear that that you have not on strength of those stations you do hear that they may be put on the speaker without L.F. amplification. Overloading must, however, be considered when strong signals of this type are available.

A Decimal Point

"Could you please let me know whether a .0005 tuning con-denser is the same as a 0.0005 condenser. The set I am building has a number of condensers marked 0.0005, 0.0003, and 0.003, etc."-G. P. (Birmingham.)

THE cipher proceeding the decimal point is often employed in order to enable the decimal point to be more clearly identi-fied. If this first figure is not employed there is sometimes a risk of the decimal point being mistaken for a mark on the paper, etc., but the 0 preceding it makes it quite clear. Therefore 0.0005 is exactly the same as :0005.

Microphone Transformer

"I have a microphone which I was told required a transformer. Not having seen any data concerning this type of transformer, I tried an ordinary 3 to 1 L.F. transformer. Results were very disappointing. Are there any special details to remember with transformers for this purpose ? "-V. W. (Bath.)

THE L.F. transformer is of no use, although you do not state what type of microphone you were using. Generally, the transformer used in conjunction with a carbon type of microphone has a very high ratio, somewhere in the neighbourhood of 75 to 1. You should endeavour to ascertain some details con-cerning the particular microphone, in order to obtain a satis-factory type of transformer for use with it. See the December, 1941; issue.

Replacing Condenser

"I have a mains set which has developed a fault that I have now "I have a mains set which has developed a fault that I have now identified as being due to a fractured electrolytic smoothing con-denser on the H.T. side. This is a small waxed cardboard box with three leads coming from it, two red and one black, and the wax has run and I have tested the H.T. and find none there. I am going to replace the condenser but I am not certain which lead is which, and as I want to use a wet electrolytic condenser in place of these cardboard items I should like to know how to identify the leads."—H. Y. (Slough.)

THE condenser block obviously consists of two condensers, and THE condenser block obviously consists of two condensers, and these are joined together inside the casing with a single lead from the two negative poles. This lead is the black one, Therefore, when replacing this unit by two ordinary electro-lytics they should be mounted so that the case of each is in contact with the chassis (or the H.T.- line) and the centre terminal points on the condensers should be joined to the points to which the red leads are now connected.



PRACTICAL WIRELESS

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ELECTRO-MAGNETIC COUNTERS. Re-sistance 500 ohms. From 1-12,990, size 44" x 2" x 14". Ex-G.P.O. 9/6 each. 1-1099, size 49 x 1/y x 14", 5/6. VARIABLE CONDENSERS. 2 and 3-gaug. Die. cast frames in first-class condition, 6/9 and 7/3 each. YAXLEY type switches, 4 bank 4-way, 7/6; 4 bank 4-way, 8/6; 3 bank 4-way, 7/6; 4 bank 4-way, 8/6; 3 bank 4-way, 7/6; 4 bank 4-way, 8/6; 3 bank 4-way, 7/8; 4 bank 4-way, 8/6; 3 bank 4-way, 7/8; 5 mail relays for operation on 2 v. D.C. with 6-way make and break switches. Brand new, 5/- cach. TRIMMERS. Twin trimmers on ceramic base, new, to clear, 6d. each, 5/- doz. COIL FORMERS. Cardboard and Paxolin. Assorted size, 2/9 doz.

Coil FORMERS, Cardineard and Paxolin. Assorted sizes 2/9 doz. VALVES. Stock of Mullard TSP 4's, HVR 2's and TSE 4's, 17/6 each. TELEVISION diole valves. Unused Mullard EA50, 60 mm. x 12 mm. overall,

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Continued top of column 3.)

ELECTRADIX BARGAINS

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