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|  |  |
|  |  |
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| WMmers |  |
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| - - - - - |  |
| B 44 LL $\frac{3}{16}^{-}-4.75 \mathrm{~mm}$ | screwdriver face |

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OOutput： 50 W TFF， 25 W per chamnet；34W conthuuous zine wavie Rerponse： $\pm 1 \mathrm{~d} B, 18$ to o，000 Ez O Output Imppodances： 4 throuph 16 ohnsz， at 60 Hz and 7000 Hz mixed 4.1 of under $1 \%$ at input gensitititr：Mz mixed ati at rated power and Auxillary， $500 \mathrm{~m} \overline{\mathrm{~F}}$ O Power Requirement： 230 ． $250 \mathrm{~V}, 50 \mathrm{~Hz}$ a．c．
Recommended List Price e34．19．6： 24 GNS． Assembled 29 GNB．Teak Cwe Es eztris

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 50 Hz a．f．； 16 W Covers 200 to 400 zEz and B 50 kHz to 30 MHz in 5 Band－1witchod angey Buperhaterokye Gircuif inciuder controied zogenerative joc．skege A．F．C．reduces fading and bleating；A．N．L．cuta noise to a minimum Aerisl Trimmer offectively peaky incoming signals for beat reception Inuminated＂S＂


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50,250 . 1,000 V. A.C. Volts: $50,250,1,000$ V. A.C. Volts:
$0-2.5,10,50,250,1,000$ V. D.C. $0-2.5,10,50,250,1,000$ V. D.C
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. 500 mA . Resistance: $0-5 \mathrm{~K} .50 \mathrm{~K}$ $0-500 \mathrm{~K} .5$ MEGS. Decibels P. \& P. $3 / 6$.

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$50,250,500,1,000 \mathrm{~V}$,
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81.5 bB e 88.17 .6 . P. \& P. $3 / 6$. $T E-80020,000 \Omega / V O L T$ GIANT MULTIMETER Mirror soale and overload meter 2 colour scale $0 / 2.5$ / $10 / 250 / 1.000 / 5000$ v. $0 / 25 / 125 / 10 / 50 / 250 / 1,000$; $5,000 \quad$ r. D.C. $0 / 50 \mathrm{uA} / 110$ $02 \mathrm{~K} / 200 \mathrm{~K} / 20$ MEG. OHM. 815.0.0. P. \& P. $5 /-$.
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Volts: .125, .25, 1.25, 5, Volts: $125,25,1.20,5$,
$10,25,50,125,250,500$, 1,000 V. A.C. Volts: 1.5 ,
$3.5,10,25,50,125,250$, $500,1,000 \mathrm{~V}$ D.C. Cur$25,50,250,500 \mathrm{~mA}, 5,10 \mathrm{amp}$. Resistance. $=20$ to +85 dh . 12100 Decibels:
P. P. $3 / 6$.

$$
\begin{aligned}
& \text { D. Current: } 10,100 u \mathrm{~A}, \\
& 10,100,500 \mathrm{~mA}, 2.5,10 \\
& \text { amp, Resistance: } 1 \mathrm{~K}, 10 \mathrm{~K}
\end{aligned}
$$ 100 MmG . Resistance: $1 \mathrm{~K}, 10 \mathrm{~K}$. $100 \mathrm{~K}, 10 \mathrm{MEG}$, tic Case with Carrying Handle. Size 7 tin tic Case with Carrying Handle. Size $7 \frac{1}{8} i n$

$\frac{\text { P. P. } 3 / 6 .}{\text { AVOMETER MOVEMENTS }}$


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ROGERS Ravensbourn
ROGERS Ravensbrook．
ROGERS Ravensbrook
（cased）${ }^{\text {SINCLAR }} 2000 .$.
SINCLAIR Project 60
TELETON GT 101 ．
All above tuners are complete with MPX Stereo All above Decoder except where starred．

## TUNER／AMPLIFIERS

AKA） 6500
ARENA R500
ARENA 2600 Stereo AMM／FM
ARENA 2700 Stereo
ARENA T1500
ARMSTRONG M88 Decoder
ARMSTRONG 525
ARMSTRONG 520
GOODMANS 300
PHILIPS RH 790
 PIONEER SX770 AM／FM ．． PIONEER SX990 AM／FM －TELETON F2000

＊TELETON R4200
TELETON TFS50

5315
4117 5315
4117
910 4499
$\stackrel{\rightharpoonup}{\infty}$
 － SHURE M32－3 SHURE M44－7
SHURE M44C SHURE M44C SHURE M55E SHURE M75－6 SHURE M 75E－95 SHURE V15－11 SHURE V15－11
SHURE MTE SHURE M75E
SHURE M75E／D 19
 Starred cartridges above are ceramic．All others are magnetic．


## 保

GARRARD SP25，fully wired with Goldring $G 600$ Magnetic Cartrige，Complete wit ARENA SP25，with base and DUAL 1219 transcription GARRARD 408 GARRARD AP75

GARRARD SL75
GARRARD SL95 B
GARRARD SL72 B．．．．．．．．
Base and Cover to fit GAR－

GOLDRING 705 P
GOLDRING GL69 Mik，ii
GOLDG G175
OLDRING GL75 P
and 75 P Covers for 69P
with plinth cover and G800
E Cartridge SHURE M32E SHURE M44－5
and 3500.
 WHARFEDALE 100.1 $\qquad$
$\qquad$ $\begin{array}{rrrrrr}63 & 5 & 0 & 49 & 19 & 6 \\ 131 & 5 & 0 & 105 & 0 \\ 120 & 0 & 0 & \end{array}$ TELETON CR55 All others take both ceramic and macnetic cartriciges All others take both（anere decoder is extra）all the Except for Armstrong（wheredecoder is extra）alt MPX
above Tuner Amplifiers are complete with MPX Stereo Decoder．

CARTRIDGES
GOLDRING 800 Cartridge． GOLDRING 800 H GOLDRING 800 Super E． ＊GOLDRING CS90 Stere ＊GOLDRING CS $91 /{ }^{\text {GOLDRING G850．}}$ ． ORBIT Magnetic ORTOFON SL15E
ORTOFON $2 \times 15 K$ Trans－ former． SUREM3DM SHURE M31E －

COMET for after－sales service THROUGHOUT THE U．K． Pictured．Service Dept at Clough Rd，Hull also at Leeds，Goole， Wakefield，Doncaster， and Bridlington
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GOODMANS 3025.
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McDONALD 610.0 Ö．．．．．．
Base to fit McDONALD

Base to fit MCDONALD turn－
table
Cover to fitMcDONALO turn－

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PHILIPS GA146
PHILIPS 217 ．．．．．．．．．．．．．
PHILIPS 202 Electronic

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PIONEERPLX 25 cover
THORENS，TD125．．．． THORENS TD 150A M
THORENS TD 125 AB THORENS TD $150 A B$ Mk． $11 .$.
THORENS TD TX11 Cover．．

## PICKUP ARMS

GOLDRING Lenco L

GOLDRING Lenco L | GOLDRING Lenco L69．．．．．．．． | 12 | 6 | 5 | 9 | 7 | 0 | 0 |
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SPEAKERS
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B．\＆W．Mod
B \＆WM3
B\＆WDM COitor 120 GELESTION Ditton 15 GOODMANS Minster GOODMANS Malesta GOODMANS Maxim． 3 GOODMANS Mezzo
GOODMANS Magnum K2 GOODMANS Magnum
GOODMANS 3005 （pair） KEF Celeste
KEF Concord
KEF Concert
KEF Cresta
KELETRON KN654／3 Speaker system（pair）
KELETRON KNB24／3 speaker system（pair）
KELETRON KN104／3 KELEAKR system 123 Speaker system ${ }^{\text {SELETRON }}$ KN120／4 LEAK 300 ．． LEAK 200
LOWTHER Acousta（with
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Rosedale
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Unit 3 Speaker K
Unit 4 Speaker Kit

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## TOPIC ロF THE MDNTH

## Designer's Trophy 1970

THE PROJECT AUTUMN competition has closed, the entries have been read, the equipment examined-and the judges have retired feeling slightly type-happy. As a result of all this, it is with pleasure that we can now announce that the Practical Wireless Designer's Trophy for 1970 has been awarded to Caleb R. Bradley for his winning article, "A High Impedance Voltmeter". The formal presentation of the silver cup will be made at the Practical Wireless/Mullard Film Show at Caxton Hall on Friday, March 5th.

Other entries which were selected by the judges as deserving special mention include the following (not given in order of merit): "Transistorised Oscilloscope" by J. R. Ransome, "Stereo Decoder" by A. Barber, "Language Laboratory" by N. Howie, "Enlarger Timer" by D. Smith, "Punchcard Operated Safe" by A. Birds, "Oscilloscope Trigger Unit" by P. Rouse and "A Novel Door Alarm" by D. Cross. The winning entry will be published in the April issue and others will appear in future issues.

So much for the successful competitors, but what of the losers? Entries were judged on a combination of factors, including originality, construction, presentation, technical accuracy and so forth. Thus it is not, perhaps, surprising that many articles were returned due to non-originality, poorly written text (in the literal sense), badly designed circuits, unnecessary complexity. A few entries featured circuitry which could never work properly (some of these, however, were nicely built!) but the main reason for failure seemed to be inability to grasp the type of article. and the style required. Authors are free to develop a theme in their own way, but certain conventions should be followed (which can be determined by studying samples from recent issues). To help those wishing to try again, we will be publishing shortly some hints on how to set about writing for P.W.

Getting back to Project Autumn, readers will note that the professional regulars stood aside from the competition, thus leaving it wide open to anyone interested enough to put in the effort. Congratulations to the winner and to the runners-up in the 1970 contest and our thanks to everyone who entered into the spirit of the event and sent something along. And remember-now is the time to start thinking about your entry for the 1971 Trophy, details of which will be given in the near future.
W. N. STEVENS-Editor.

## MARCH ISSUE WILL BE PUBLISHED ON FEBRUARY 5

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## WEWS... MEWS... NEWS...

## Compact Unit Audio



## Bristol reader wins Hi-Fi Competition

Remember the Hi-Fi Competition which appeared in our November issue? Entrants were invited to put in order of importance eight features applying to the Decca 3000 Series audio unit system.

All entries were examined and it was decided that the best entry was that submitted by Mr. A. R. Dyke, of Bristol, whose order was: 1-J; 2-K; 3-L; 4-E; 5-A; 6-D; 7-B; 8-C.

Congratulations to Mr. Dyke who becomes the winner of a Deccasound 3000 Series Hi Fi system, worth over $£ 200$.

Our thanks, too, to all readers who joined in the competition and helped to make it such a great success.

## Mini Power Supplies

Bentron Electronics have developed a new series of miniature power supplies.

The maximum input voltage is, depending upon the model, 40 volts to 70 volts and the output voltage 4 volts to 60 volts. All parts are absolutely protected against overload and each unit is completely short circuit proof. Encapsulation by epoxy resin ensures full protection against humidity and mechanical influence which results in unusually high reliability. Further information: Rastra Electronics Ltd., 275 King Street, London, W.6.

## The Transensor

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The makers claim that this is the only recorder/sychroniser in its price range that allows one to erase the sync pulses independently of the commentary on both sides of the cassette tape.

The miniature amplifier in the Transensor is powered by a PP3 9 V battery and when the battery current is below a working strength, a warning tone can be heard on the recording position. This prevents the synchroniser from being used until a new battery is put in. The switch of the Transensor is situated in the 4-pin DIN plug connecting lead, so that the synchroniser will be in the "on" position automatically when used in conjunction with the projector and in the "off" position when the plug is removed. Price is $£ 43$ which includes recorder/Transensor, microphone, C60 cassette, carrying case, lead for connecting Transensor to projector, 2 separate manuals and 1 year guarantee. Audio Visual Picture Enterprises, 17 Abercorn Place, London, N.W.8. Phone 01-328 1461.

The Apollo Unit Audio Hi-Fi system employs a Garrard 2025 TC deck and Garrard G.C. 536 cartridge. Amplifier output is 5 W per channel, frequency response 40 $20,000 \mathrm{~Hz}$ and controls are volume, bass, treble and balance. Speakers are in polished teak and comprise a 6 in. twin cone bass unit and a $2^{1}{ }_{2}$ in. parasitic tweeter with LC crossover. The suppliers stress that this unit is $100 \%$ made in England and supply a 12 months guarantee. The plinth, cover and amplifier unit measures $141_{2} \times$ $12^{3} \times 7^{1}{ }_{2}$ in. and the speakers measure $14^{1}{ }_{2} \times 8 \times 6^{1}{ }_{2} \mathrm{in}$. The price is 36 gns . and postage, packing and insurance come to 50 s . Audio Supplies (PW) 50 Stamford Hill, London, N. 16.

## £20 Prize Winner

The winner of the $£ 20$ prize in the competition run in conjunction with Public Opinion Surveys questionnaire is: Mr. H. C. Anthony, 25 Edwalton Avenue, Peterborough.

## Trouble getting P.W.?

Some readers report having difficulty in obtaining their copies of Practical Wireless. We strongly recommend that if this is the case, a regular order should be placed with a local newsagent. If difficulty is still encountered, please write to me direct, giving all the relevant information.Editor.


## MEWS... NEWS... NEWS...

## More Room

A useful working surface for those with small shacks or just a corner of the kitchen, is provided by the Foldette-a folding table. Fixed to the wall by a simple: do-it-yourself system, it can be pulled down with one easy movement, to provide 8 square feet of peacerestoring extra space.

The Foldette takes up no valuable floor space whatsoever. It has no legs to impede the endless to-ing and fro-ing that goes on, and when finished with, it can be pushed up again, with just one hand, to lie flat, protruding only $2^{3}{ }_{4} \mathrm{in}$. from the wall. It has a specially designed balance mechanism engineered to return the table to the wall automatically if it is pulled down less than 30 degrees. If it is pulled down more than that it stays put-and doesn't descend on to your head just as you bend to pick up a dropped screw. Gentle downward pressure brings it into the positive, ready for use position where it can accept a weight of up to 500 lb .
A choice of three colours for the wipe-clean plastic laminate top fits the table into most decorative schemes. White, Blue, and Yellow, are available, with two patterns-Grey Tweed and Teak Wood Grain. The strong, steel frame is finished in black stel-vetite-a coating which gives a pleasant grained effect to the metal. The table costs approximately $£ 25$, with the mechanism guaranteed for 12 months. The Foldette is supplied complete with everything needed for fixing, including clear, easy-to-follow instructions. Polydesigns Limited, Coronation Road, High Wycombe, Bucks.

## Baker Speakers

A new model-the "Regent" has been introduced by Baker Reproducers Ltd. and sells at $£ 7$. The D.W.1. Laboratory Loudspeakers 12in. model now has a double central metallic ellipse.

Copies of the leaflet giving details of Baker speakers may be obtained from Baker Reproducers Limited, Bensham Manor Road Passage, Thornton Heath, Surrey.


The Foldette table shown in folded and open positions.


THE E.H.T. Voltmeter, to be described, may be built in two versions, either as a simple single range $0-25$ kilovolt meter, consisting of high resistance series resistor and a $0-50$ microamp meter, or as a sophisticated multi-range instrument which will have many uses in the electronics field, as well as in television servicing. The multi-range version has ranges of $2 \cdot 5 \mathrm{kV}, 5 \mathrm{kV}, 10 \mathrm{kV}$ and 25 kV , with a constant input resistance of $500 \mathrm{M} \Omega$ on all ranges.

## MULTI-RANGE

## 

## VOLTMETER

 J. THORNTON LAWRENCE
## Single-range Version

The basic circuit of the single range E.H.T. Voltmeter is shown in Fig. 1. The internal resistance of the meter is negligible compared with the series resistor Rl and so can be neglected. The deflection of the meter is determined by the input voltage and the series resistor, giving a full scale reading of 25 kV .

$$
\operatorname{Im}(50 \text { microamps })=\frac{\operatorname{Vin}(25 \mathrm{kV})}{\mathrm{R} 1(500 \mathrm{M} \Omega)}
$$

The series resistor is identical to that used in the multi-range version and the construction of it is described later.


Fig. 1. Basic circuit of single-range EHT voltmeter.

## Multi-range Version

The multi-range version of the E.H.T. Voltmeter employs an integrated circuit operational amplifier of the '709' type, operating as a d.c. amplifier and driving an indicating meter, as shown in Fig. 2.

In the circuit, the operational amplifier has negative voltage feedback connected in shunt with the input signal. The ' 709 ' type of amplifier has an 'open loop' gain in excess of 20,000 , and as the demanded gain for the circuit, on the most sensitive range, is only 10 , the overall voltage gain of the amplifier circuit, for all practical purposes, is determined entirely by the ratio of the feedback resistor

Rf to the series resistor R1. The output voltage can be expressed as:

$$
\text { Vout }=\operatorname{Vin} \times \frac{\mathbf{R f}}{\mathbf{R 1}}
$$

The meter deflection depends on the output voltage of the amplifier and the meter series resistor Rm (which includes the resistance of the meter), and may be expressed as:

$$
\operatorname{Im}=\frac{\text { Vout }}{\mathrm{Rm}}
$$

The meter deflection for any input voltage can therefore be expressed as:

$$
\operatorname{Im}=\frac{\operatorname{Vin} \times \mathbf{R f}}{\mathbf{R 1} \times \mathbf{R} \mathbf{m}}
$$

To change the voltage range, it is necessary to change the value of one of the resistors in the circuit, and the most convenient one to change is Rf.


Fig. 2. Basic circuit of multi-range EHT voltmeter using an ic. operational amplifier.
Working out the values of $R f$ required to give the desired voltage ranges, with $\mathrm{RI}=500 \mathrm{M} \Omega, \mathrm{Rm}=100 \mathrm{k} \Omega$ and $\mathrm{Im}=50$ microamps, we get the following results:

| Vin | Rf for full scale reading |
| :---: | :---: |
| 25 kV | $100 \mathrm{k} \Omega$ |
| 10 kV | $250 \mathrm{k} \Omega$ |
| 5 kV | $500 \mathrm{k} \Omega$ |
| 2.5 kV | $1 \mathrm{M} \Omega$ |



Fig. 3. Complete circuit of the multi-range voltmeter. Switch contacts and knob rotation shown from rear.

In the practical circuit, the meter series resistor, Rm , actually consists of a fixed resistor of $82 \mathrm{k} \Omega$, a pre-set variable resistor of $30 \mathrm{k} \Omega$ and the internal resistance of the meter. The variable resistor is adjusted to give the required total value of $100 \mathrm{k} \Omega$. Should R1 not be exactly 500 M , due to resistor tolerances, the value of Rm can be adjusted slightly by this variable resistor to compensate for these inaccuracies and thus give the correct full scale reading.

## Circuit

The full circuit of the Multi-range E.H.T. Voltmeter is shown in Fig. 3. The e.h.t. voltage to be measured is connected to the input terminal and the current through R1 is passed through S1a to the inverting input of the operational amplifier, pin 4. The output of the amplifier, pin 10, is fed back to the input through the selected feed-back resistor, R5 to R8, and S1b.
The output of the amplifier is also fed through R11, VR2, S2b, S2c and the meter, M1. Switch S2 performs several functions, section S2b and S2c reverse the connections to the meter for positive and negative e.h.t. inputs, and short circuits the meter when the instrument is switched off. Sections S2a and S2d connect the positive and negative supply rails to the positive and negative batteries when the instrument is in use.

An internal battery check position is provided on Sla by switching to R3 to check the positive supply battery and to R4 to check the negative supply battery. The range of the meter in these positions is 10 volts full scale.

The operational amplifier has a small input current which would cause an off-set voltage at the output. A compensating current is provided by VR1 and R9 to bring the output of the amplifier to zero. This control is the "Zero" control and is adjusted on the most sensitive, 2.5 kV , range for zero indication on the meter.
Capacitors C1 and C2 provide decoupling for the amplifier and C3, R10 and C4 are compensation components to ensure stability of the amplifier when operated with considerable negative feedback.

Resistor R2 is included at the bottom of R1 to prevent voltage building up across Sla if the switch was in the battery check position, with e.h.t. connected to R1. R2 does not affect the accuracy of measurement as it shunts the input of the amplifier, which is very low resistance due to the large amount of negative feedback being used.

## E.H.T. Resistor

The heart of the E.H.T. Voltmeter is the chain of voltage dropping resistors which have to be capable of being connected to 25 kV . Because of the high voltage involved, it is necessary to employ a special method of construction that will keep the voltage across each individual resistor to a satisfactory value and to allow adequate insulation between adjoining resistors in the chain and to surrounding objects.


Photograph of the partially assembled e.h.t. resistor chain.

This is done by mounting fifty $10 \mathrm{M} \Omega 1_{2}$ watt resistors on a framework by two perforated S.R.B.P. (Synthetic Resin Bonded Paper) boards which are housed in a plastic beaker, Fig. 4.


Fig. 4. Assembly details of the e.f.t. resistor in its protective container.
The $10 \mathrm{M} \Omega$ resistors are fitted in a spiral fashion by starting at one end and moving one line of holes ( $0 \cdot 1$ inch) further as each resistor is fitted, so that between each resistor and the one directly below it, there are three blank holes, as shown in the drawing. It is important to make clean soldered joints, with no spikes of solder or wire sticking out as at very high voltages any sharp points will cause corona breakdown.
To reduce any tendency for corona discharge to take place, from the top input terminal of the resistor assembly, it is fitted with a stress reducing par-tial-sphere consisting of a round edged tobacco tin.

## Instrument case

The metal for the top cover is cut, marked out, drilled and folded as shown. The position of the bends have been arrived at empirically and it is advisable to check the position of the markings by physically comparing these with the lower part of the case before actually making the bend.

When the cover is fitting snugly on the lower part of the case, the fixing screw holes can be marked through and then drilled for self-tapping screws or drilled and tapped 6BA. The case may be sprayed with car touch-up enamel or covered with 'Fablon' or 'Contact' material and lettered with 'Letraset.'

## Meter and Amplifier

The amplifier circuit is built on a $0 \cdot 1 \mathrm{in}$. $\times 0 \cdot 1 \mathrm{in}$. matrix Veroboard panel, size 3 in $\times 2 \cdot 6$ in., which is

## components list

Resistors :


Integrated Circuit : Motorola MC1709C (Bi-Pak BP709)
I.C. socket (Bi-Pak TSO14)

## Switches:

S1 2-pole 6-way midget wafer
S2 4-pole 3-way midget wafer
Miscellaneous:
Meter M1, $50 \mu$ A f.s.d. (Henrys MRA-65B),
Veroboard (1) $0.1 \times 0.1 \mathrm{in}$. matrix, $3 \times 2.6 \mathrm{in}$.
Matrix boards (2) (Radiospares Type 187).
Knobs ( 1 Type WK, 1 Type PK with trims, Radiospares).
Batteries (2) PP9, and connectors.
Polythene beaker (Tudor Rose, Woolworths).
Earth terminal, plastic feet, high-tension cable.
mounted directly onto the meter terminals. The layout of the plain side of the Veroboard panel and the wiring connections are shown in Fig. 5. A break in the copper strip, on the reverse side, is shown as an ' X ' on the drawing.

It is necessary to use the $0 \cdot 1 \mathrm{in}$. matrix Veroboard because the Integrated Circuit holder used has a 0.1 in . pin spacing. As there is only 0.05 in . between the copper strips, it is important to use a small soldering iron bit and to avoid using excess solder when fixing the components as this may result in a short circuit between adjacent strips.

## Set Zero and Calibration Controls

The ZERO and CAL controls are of the pre-set type and the knobs protrude through the back of the instrument. To avoid any possibility of accidental misadjustment, the CAL control is set in from the panel by using spacers and long screws so that the knob is only just showing, whereas the ZERO control knob protrudes the normal amount.

## Range and Polarity Switches

The range switch is a 2 -pole 6 -way midget wafer switch and is mounted in the upper hole on the front panel. The switch positions, reading from left to right, are $-,+, 25,10,5$ and $2 \cdot 5$. The first two positions enable the negative and positive batteries to be checked and the other positions indicate the full scale reading in kilovolts.

The polarity switch is a 4 -pole 3 -way midget wafer switch and is mounted below the range switch. The positions are - , OFF and + .

## Connecting Leads

The earth lead from the Voltmeter should be of a reasonable size, say $23 / \cdot 0076$, and should terminate


Marking out and drilling details for the top of the instrument case.
in a reliable crocodile clip, so there is no danger of the lead fracturing or becoming un-clipped. The e.h.t. connecting lead should be of special high tension cable, such as Radiospares 'E.H.T.' cable, which is suitable for operation up to 18 kV .

For voltages higher than this it is possible to remove the braiding from co-ax cables, to obtain a conductor with the required insulation. The inner conductor and insulation on Uniradio 43 is good to 21 kV and Uniradio 67 to 38 kV .

Some kind of probe tip arrangement is useful when working on voltages up to about 5 kV , but a simple hook at the end of the probe is probably the best arrangement for general use as it can be hooked on and left without being held. The probe used with the prototype was made from a giant ballpen case.


The E.H.T. Voltmeter should be carefully connected to the circuit under test by first connecting the earth terminal of the voltmeter to the chassis of the equipment. In the case of t.v. receivers, all the usual precautions regarding mains connected chassis should be observed.

It is most important to exercise the greatest care when working on the e.h.t. circuits of electronic equipment where these are derived directly from a 50 Hz mains transformer. Always stand on a rubber or plastic mat, work using one hand only and keep the other in your pocket and never stick your head near the e.h.t. voltage or inside the equipment. After switching off, ensure that all the e.h.t. capacitors are discharged by shorting them out with an earthing lead.

## Calibration

As with any transistor equipment, the wiring should be double checked, particularly the polarity of the battery supplies, before switching on.

Flg. 5. Circuit board component layout.

1. With the Polarity switch at OFF, check the mechanical zero of the meter.
2. Set the Range switch to $2 \cdot 5 \mathrm{kV}$.
3. Set the Polarity switch to +.
4. Adjust the amplifier ZERO control for zero indication.
5. Switch the Polarity switch to - and check that the meter is still at zero.
6. Connect the meter to an accurate supply of $2 \cdot 5 \mathrm{kV}$ and set the CAL control to give full scale reading. The accuracy on the other ranges should be within about $3 \%$.


Practical wiring layout of the EHT multl-range'voltmeter.
If no accurate e.h.t. supply is available the meter may be calibrated as follows:

Fit two new PP9 batteries.
Complete operation 1 to 5.
7. Switch the Range switch to + .
8. Switch the Polarity switch to + .
9. Adjust the CAL control for an indication of 9.6 volts, ( 48 on the $0-50$ scale).
10. Switch the Range and Polarity switches to - and check that the voltage reading is the same.
11. If there is any difference, adjust ZERO control so that $9 \cdot 6$ volts, ( 48 on the scale), is midway between the + and - indications.

The calibration of the e.h.t. ranges should now be accurate to within about $7 \%$ of the full scale reading of the range in use.

## THE

MWN columi


ASIATICS have been prominent on the medium waves this winter. On some parts of the band they appear in the spaces between European channels and are audible in the evenings before local stations close down for the night. Hopei 940 kHz which lies between AFN 935 kHz and Toulouse 944 kHz , has been picked out several times by the writer at 2200 hrs GMT. The programmes are unmistakably Chinese. Calcutta 1130 kHz can be found on the l.f. side of the burble on 1133 kHz , it signs off at 1700 hrs but has been logged again at 2210 hrs after sign on. Three Japanese stations reported recently to Medium Wave News are JOIB Sapporo on 750 kHz at 2115 hrs GMT; JOBB Osaka 830 kHz at 2030 hrs ; JOLB Fukyoka 870 kHz at 2105 hrs while Chinese have been reported between 2100 hrs and 2230 hrs GMT on $660,740,770,860,920,1020,1280$, 1290 kHz . KSBU 1360 kHz in the Ryuku Islands (Okinawa) now signs on at 2000 hrs and has been logged at 2130 hrs . Taiwan (Formosa) has been heard at 2000 hrs from BED96 900 kHz and at 2100 hrs from BED97 1100 kHz . The VOA operates a 1000 kW Voice of Free Asia in Thailand on 1580 kHz which was heard recently in Holland at 2200 hrs and by the writer at 2240 hrs . Bagdad Iraq is a consistent signal during the evening on 760 kHz while from the Persian Gulf there is Kuwait 1345 kHz and the BBC relay on Masirah Island on 1410 kHz , both being logged at 2100 hrs . Omdurman Sudan on 960 kHz is on the air until 2200 hr s with Arabic programming but should not be confused with the stronger signal from Tunis on 962 kHz .

There is an Asiatic station on 200 kHz on the long waves which has been heard after the close down of BBC-2 at 0110 hrs GMT, with music that is neither oriental nor arab in style. The language is unknown to the writer but is probably Turkmenian, since Ashkhabad, which is located north of Iran and east of the Caspian Sea, is listed on this frequency. This station has been heard several times and on one occasion was strong enough to produce a beat on the $S$ meter before Droitwich signed off. Prague on 272 kHz closes down at 0100 hrs . GMT with multilingual announcements, to clear the channel for Novosibirsk in Siberia, which has been heard with news in Russian.

The two oldest broadcasters in the United StatesKDKA and WBZ-which were featured recently in the Going Back series, are still on the air and are often logged by DXers in this country. KDKA Pittsburg is on 1020 kHz and WBZ Boston is on 1030 kHz . Listen after midnight when the path is open, and both will be heard. The third oldest is WBT Charlotte North Carolina now on 1110 kHz , which is usually logged together with CBD Saint John, New Brunswick on the same channel. Other U.S. stations in the centre of the band are WINS 1010 kHz in New York City; WHN 1050 also in NYC; WBAL 1090 in Baltimore. From Canada there are CKBW 1000 kHz in Bridgewater Nova Scotia; CJRP 1060 in Quebec City (in French) and CBA 1070 Moncton N.B.

Charles Molloy

# RMATEUR雷 Banos COMWUNICAIIOWII RECEIUER 

## RF Amplifier/First Converter

In order to reduce the complexity of the switching, the tuned circuits are arranged round a 4 -wafer, single pole, 8 -way switch, Sla-d. Any compromise is more than made up in efficiency, particularly as there is only one switch contact per circuit instead of the usual two or three, plus a bonus of considerably simplified wiring. The circuit is shown in Fig. 11.
Tr1 is used in the common collector mode, which provides a sufficiently high input impedance (about $100 \mathrm{k} \Omega$ ) directly across the tuned circuit.
The 7 MHz coil L 2 is permanently in circuit, a capacitor C 2 being switched across it in the $3 \cdot 5 \mathrm{MHz}$ position to resonate the coil in this band, while on

of low noise, the oscillator injection is at a rather low level (about 100 mV ). The output from the mixer is via the screened lead to Ll in the second converter stage.

Probably the nearest that one can get to stability approaching that of a crystal (in an oscillator circuit) is the series tuned Colpits circuit. This has a number of good points, not least being the fact that small changes in inductance have little effect on oscillator frequency, which is as well, as the band switch has to be part of the inductance. In fact, due to being in the low impedance end of the coils, the switch has no effect on oscillator frequency. It is possible to zero beat a signal on one band and then switch

the h.f. bands another coil is switched in parallel to resonate in the required band.
The series resistor R10 in the base of $\operatorname{Tr} 2$ deliberately holds down the gain which would otherwise be excessive, and also reduces the loading of Tr 1 , at the same time helping to keep the gain constant on all bands, as the load is now almost purely resistive.

The same method of switching is used in the collector circuit of $\operatorname{Tr} 2$. A low impedance link carries the signal to the mixer $\operatorname{Tr} 3$ via C8. In the interests


Fig. 11: Circuit of the r.f. amplifier and first converter stages with details of the bandswitch connections.
around, returning to this band, when the signal will still be in zero beat.

As a point of interest, it should be possible to use this converter with any receiver covering 1.5$2 \cdot 0 \mathrm{MHz}$ (or even $1 \cdot 0 \cdot-\cdot 5 \mathrm{MHz}$ ) by taking $\operatorname{Tr} 3$ collector to the 9 V line via an r.f. choke or a $2 \cdot 2 \mathrm{k} \Omega$ resistor, and coupling via a $0.01 \mu \mathrm{~F}$ capacitor and screened lead to the receiver aerial and earth sockets. Test voltages are given in Table 4.

## Construction

The unit is built up in a box chassis, fabricated from tinned plate with aluminium end panels, each stage being screened from its neighbour by a partition screen, also of tinned plate.

The chassis measures $6 \times 4 \times 1 \frac{3}{4} \mathrm{in}$ and the assembly details are shown in Fig. 12. Partitions and end plates should be cut very slightly undersize. Drilling is completed before the pieces are assembled but note that the two screws holding the front of the gang also secure that side of the front plate. Take care that all the holes for the switch wafers are aligned correctly, any undue stress on the switch sections will cause excessive wear and erratic operation later on.

TABLE 4

|  | Collector | Base | Emitter |
| :--- | :---: | :---: | :---: |
| Tr1 | 8.9 | 1.0 | 0.4 |
| $T r 2$ | 8.8 | 1.4 | 0.9 |
| $T r 3$ | 9.0 | 0.9 | 0.3 |
| $T r 4$ | 4.5 | 2.0 | 1.4 |
| Measured with $20 \mathrm{k} \Omega / \mathrm{V}$ meter |  |  |  |

When all the parts of the box are ready, the front plate and gang are fitted first, then the switch shaft, indexing mechanism, and first switch wafer. The mounting position of the wafer is as Fig. 11. Mark the positions of the three partitions as these are fitted only after the wiring-in of the components in each previous section has been completed.

Next wire up the first stage as in Fig. 12. The coil formers are secured with impact adhesive and earth returns are soldered direct to the chassis.


Fig. 12: Wiring details of the r.f. amplifier and first converter stages, Winding information for the inductors is given in the adjacent table.

The first partition, with the second wafer secured, is now soldered into position, and wiring of the second stage completed. The procedure is repeated for the third and fourth stages.

## Alignment

This will be made easier, if the r.f. coils are first roughly pre-tuned. To do this, solder a diode (OA70 or similar) to the live side of VCla and connect a meter from this to chassis. Set the gang to midposition and inject a $3 \cdot 8 \mathrm{MHz}$ signal into Ll , switching to the 80 m band. Adjust the core of L 2 for maximum reading on the meter. The reading will depend on the output available from the signal generator, but can be expected to be half a volt or so. If more, the output: from the generator should be reduced; if less, use the 50 or 100 microamp range of the meter. An indication of maximum reading is all that is required. Repeat the procedure on the 14 and 21 MHz bands.
The coil L 2 is common for both 7 and 3.5 MHz bands and L 4 for $21^{\prime}$ and 28 MHz bands, the swing of VCla being sufficient to accommodate the two adjacent bands in each case.
The coils in the second stage can be adjusted in the same way by connecting the diode to VClb and injecting the signal into L5.
Note that the unit is not connected to a power supply for these adjustments.

## Assembly and Final Alignment

With the aid of the cover photograph of the receiver (December 1970 issue of PW) and the below-chassis view in Part 2 (January 1971) the various units can now be assembled on to the common chassis which is $12 \times 9 \times 2$ in and made of 18 s.w.g. aluminium. The panel $12 \times 4 \frac{1}{8}$ in ( $12 \times 4 \frac{5}{8}$ in with flange) is bolted to the front to carry the tuning

## INDUCTORS

|  |  | Function |
| :---: | :---: | :---: |
| L1 | 3 turns p.v.c. insulated wire over earthy end of L2 |  |
| L2 | 50 turns 36 s.w.g. enam. close wound |  |
| L5/6 | As L1/2 | $\begin{aligned} & \text { AE 80/40 } \\ & \text { RF 80/40 } \end{aligned}$ |
| L3, L7 | 28 turns 32 s.w.g. enam. close wound | AE/RF 20 |
| L4, L8 | 15 turns 24 s.w.g. enam. close | AE/RF 20 |
|  | 65 wound 36 mid | AE/RF 15/10 |
| L9 | 65 turns 36 sw.g. enam. close wound | Osc. 80 |
| L10 | 40 turns 32 s.w.g. enam. close |  |
| L1 | wound <br> 20 turns 24 s.w.g. enam. close | Osc. 40 |
| L11 | 20 turns 24 s.w.g. enam. close wound | $\text { Osc. } / 20$ |
| L12 | 15 turns 24 s.w.g. enam. close |  |
| L13 | wound <br> 9 turns 24 s.w.g. enam. close wound | $\begin{array}{ll} \text { Osc. } & 15 \\ \text { Osc. } & 10 \end{array}$ |

Above inductors wound on Radiospares formers $\frac{1}{4} \mathrm{in}$. dia. with dust cores.

L14, 15 Each 2 turns 24 s.w.g. tinned copper wire tin. i.d. selfsupporting

Osc. 10



Tel. 01-452 0161/2/3 A. MARSHALL \& SON $\begin{gathered}\text { Send I/ For new Comprehensive Semi- } \\ \text { conductor and I.C. list (24 pages) }\end{gathered}$ Telex 21492

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scale, drive and $S$ meter. This panel is mounted with the flange forward so that the scale is recessed behind the front panel proper which can also be of aluminium with cut-outs to show the scale and $S$ meter.
The interconnections of the units can now be made and the alignment of the r.f. stage and first converter completed.

Connect the voltmeter to the a.g.c. test-point, set the main tuning control to 2.0 MHz (near minimum capacity) and "R.F. Tune" to near maximum capacity. Switch to the 80 metre band and inject a signal of $3 \cdot 5 \mathrm{MHz}$ and adjust the core of L 9 for maximum, reducing the input signal to obtain accurate tuning. Now adjust L6 and then L2 for maximum, always reducing the generator output so that the meter reading (a.g.c. voltage) does not exceed about 3 volts.
Leave the main tuning at 2 MHz and reset "R.F. Tune" to mid-position and repeat the above procedure on the remaining bands. Note that the r.f. cores will not require adjustment on the 40 metre or 10 metre bands; it is only necessary to tune "R.F. Tune" to maximum on these bands.
In each case, the oscillator is tuned to a signal injected at the low frequency end of the respective band, i.e., 7, 14, 21, 28, 28.5 and 29 MHz . In the case of the 10 metre band, L13 is adjusted first to 29 MHz , after which L14 and 15 are adjusted if necessary to 28.5 and 28 MHz respectively, by squeezing or opening the turns as required.
It will be noted that the second converter gang is at minimum capacity when tuning the l.f. end of the bands (except Top Band) and vice versa. This is because the i.f. produced (signal out) is the

## components list


difference frequency between the fixed oscillator and signal-in frequencies. Table 5 should make this clear.

TABLE 5

| Signal In | Oscillator | Signal Out |
| :---: | :---: | :---: |
| $\begin{aligned} & 3.5 \\ & 4.0 \end{aligned}$ | 5.5 | $\begin{aligned} & 2.0 \\ & 1.5 \end{aligned}$ |
| 7.0 7.5 | 9.0 | 2.0 1.5 |
| $\begin{aligned} & 14 \cdot 0 \\ & 14 \cdot 5 \end{aligned}$ | 16.0 | 2.0 1.5 |
| 21.0 21.5 | 23.0 | 2.0 1.5 |
| 28.0 28.5 | 30.0 | $\begin{aligned} & 2.0 \\ & 1.5 \end{aligned}$ |
| $\begin{aligned} & 28.5 \\ & 29.0 \end{aligned}$ | 30.5 | 2.0 1.5 |
| 29.0 29.5 | 31.0 | 2.0 1.5 |
| Frequencies in MHz |  |  |

On the h.f. bands it is possible to tune the r.f. circuits to the other side of the oscillator; for example, to 18 MHz instead of 14 MHz to produce the same i.f. so the approximately correct position for this control for each band should be marked on the front panel.

## Aerials

A separate socket is provided for Top Band, for which an ordinary end fed aerial is suitable, although a coaxial socket could be used, to feed from an aerial tuner unit.

On all other bands, the input impedance approximates to 80 ohms, and for best results the aerial must be matched to this impedance. A dipole aerial would provide the correct impedance, but it will be difficult to have a dipole for each band! The alternative is to use an end fed aerial coupled to the receiver via an a.t.u. or a multiband dipole.

## Muting

When working in conjunction with a transmitter, the receiver can be muted by inserting a switch (or break-in relay) in the 9 V supply line after the zener diode.
It will be a wise precaution to fit a pair of low capacity diodes (such as AA119) back-to-back across the first transistor input circuit. These can simply be connected across VCla (R.F. Tune) with a physically small d.c. blocking capacitor of $1,000-5,000 \mathrm{pF}$ in series with them. The purpose of the diodes being to protect the first transistor against damage from r.f. from the transmitter. If a resistor is inserted between one of these diodes and earth, a voltage will be developed across it during "transmit". This can probably be used via a d.c. amplifier, to operate a transistor switch in the 9 V supply line (before the zener in this case).

# TAKE 2® JULIAN ANDERSON 

## A series of simple transistor projects, each using less than twenty components and costing less than twenty shillings to build.

WE get all the American electronic magazines in the office and after reading through a few of them you realise that they are nutty on "Treasure Finders". Circuit after circuit crops up and the advertising pages are full of them. I would have thought that the amount of buried treasure on the other side of the Atlantic was minute compared to that in Britain but we don't seem to have caught the bug for these.

Metal locators work on a variety of principlessome are extremely complicated-but the simplest type is very simple indeed and, when used with a transistor radio, consists only of the circuit in Fig. 1.

All the circuit does is to oscillate at a frequency in the medium wave band. The coil part of the circuit is the business end of the device. When standing in free air, away from metal objects the oscillator will work at a particular frequency which can be varied by VCl. However, as soon as a metal object is moved near this coil its inductance will increase and the frequency of the oscillator will fall. The amount of this decrease will depend on the proximity and size of the metal object. By itself this is of course no good; since the frequencies are at r.f. we can't hear them. We can, however, make this signal beat with an existing signal whose frequency does not vary and we shall then be able to hear the note as a whistle.

The existing signal we are using is the carrier of a local broadcast station; if we set our oscillator very close to that of a local station we will hear a beat note which can be tuned by VC1 and we hear this over the set's loudspeaker as a whistle. With this principle you will still hear the station but the sound will be low compared to the whistle. You don't have to make any connection to the radio set, it need only be near the oscillator (within a few inches).

## The Circuit

The simple r.f. oscillator used is shown in Fig. 1. VCl can be any 500 pF variable but the cheapest type and the easiest to mount is one of the Radiospares compression trimmers; this will have a wide enough range for our purposes. Cl is the feedback capacitor and R1 the base bias resistor. L1 is the search coil and this consists of 30 turns of wire, any normal type, wound onto a former between $3^{\prime \prime}$ and $4^{\prime \prime}$ in diameter. A tap at 15 turns is necessary. A number of things can be used for the former, in the prototype it was the cardboard inner ring from a large roll of "Sellotape". A 4ft. length of planed timber $\frac{3}{4}$ " $\frac{3}{4}$ " is used as the handle and this is joined to a shorter piece of similar section wood jammed inside the cardboard former. The wires run up the handle to the rest of the circuit. Make sure that these wires are


Fig. 1 (left) The circult of the metal locator:
Fig. 2 : (right) The constructlon of the search coil.
When built the radio should be switched on and a station tuned in. The oscillator should then be switched on and VCl varied until a beat note is heard. When the coil at the other end of the handle is moved near a metal object, the frequency of the beat note will change. Getting good results is a matter of practice but it will be found that gas pipes and electrical conduit can be traced under the floor boards and pennies inside books can be detected. As far as finding treasure is concerned that is up to you but for ambitious readers remember that the Crown Jewels of King John were lost in the Wash and still haven't been found-Good Luck!

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## Circuit

This is shown in Fig. 1, and will be described briefly because the receiver is an ideal project for the beginner.

Signals are picked up by the ferrite rod aerial, of which L1 is the medium wave winding, and L2 the long wave winding. Tuning is by one section, VCl, of the ganged capacitor. When the wavechange switch S 1 is at M, L2 is shorted out, for m.w. reception. Coloured flexible leads are attached to the aerial windings for identification. L3 is the base coupling winding, supplying the mixer transistor Trl.

L4 is the oscillator coil, with emitter and collector coupling windings. It is tuned by VC2, the second
section of the ganged capacitor. For m.w. reception C5 is in circuit, and is of such a value as to allow correct tracking. That is, L4 is tuned higher in frequency than L1, throughout the band, and the difference in frequency is that of the intermediate frequency amplifier.

When S1 is in the $L$ position for long waves, C3 and the trimmer TC2 are shunted across C5 and VC2, and change the frequency coverage of L4 for long wave reception.

The first intermediate frequency transformer is

i.f.t.l. and signals pass from its secondary to the first i.f. amplifier, $\operatorname{Tr} 2$ then to i.f.t.2, a similar transformer, and $\operatorname{Tr} 3$ is the 2 nd i.f. amplifier. These transistors have emitter bias resistors R6 and R12, each with a by-pass capacitor.


Fig. 1: Circuit diagnam of receiver. Consult maker's data sheet for connections to ferrite aerial.


Fig. 2: Top of chassis layout. Colour identification of ferrite aerlal leads was that used by author.

Amplifier signals pass from i.f.t. 3 to the demodulation and a.g.c. diode D1. This diode supplies audio signals to the volume control VR1. Rectification of the carrier by the diode Dl also develops a steady d.c. potential across VR1. This is the a.g.c. bias, which reaches the base of $\operatorname{Tr} 2$ via $\operatorname{R9} 9$ and the secondary of i.f.t.l. If signals are strong, additional positive bias reaches Tr2, reducing its gain. With weak signals, there is little or no a.g.c. bias, and gain is raised. In this way output from the receiver is maintained at a fairly steady level, despite changes in the strength of signals received.

Audio signals from VR1 reach $\operatorname{Tr} 4$ through R13 and C14, and are amplified. R18 is the emitter bias
resistor, with capacitor C15, while R14 and R15 provide correct base operating conditions. The amplified signals go to the primary of the driver transformer T1.

Tr5 and Tr6 are the output transistors, and are so biased that little collector current fiows when no signals are present. The transformer Tl has a centretapped secondary. Audio signals thus drive $\operatorname{Tr} 5$ and Tr6 into conduction when the base of either swings negative, the collector currents being combined in the speaker matching transformer T 2 , the secondary of which is connected to the loudspeaker. R17 provides negative feedback, to obtain improved audio response.


Photograph of completed receiver. Compare with Fig. 2.

## Ferrite aerial

The ferrite aerial employed has a m.w. base coupling winding L3 and tapping on L2, Fig. 1. This operates with the simplest possible switching, only S1 being required for aerial and oscillator.

The ferrite rod is supported by two pieces of insulating material as in Fig. 2, or strips of paxolin held with small brackets. Thin coloured flexible leads are soldered on, to identify connections when the aerial and windings are in position. Leads can then be taken to $\mathrm{VCl}, \mathrm{S} 1$, chassis and Cl .

Other ferrite rod and slab aerials intended for use with a 208 pF tuning capacitor should be satisfactory. With this design and layout, there is no difficulty in fitting a multi-pole switch for S 1. This would allow the use of ferrite aerials having separate base couplings windings, or requiring different switching for L1/L2. The ferrite aerial must be of the type

intended for an OC44 mixer. The actual connections for it should be made as shown by the maker's leaflet.

In the same way, in a standard circuit of this kind, it is possible to use other oscillator coils and i.f.t.'s. The oscillator coil should be for 176 pF tuning with an OC44, while the i.f.t.'s are $465-470 \mathrm{kHz}$ types for use with OC45's. When fitting other types, check pin connections against the maker's data.

## Mixer-IF board

Components are placed as in Fig. 3, the positions of the clearance holes for the oscillator coil and i.f.t.'s being most easily found by pressing the pins on paper, then holding the paper on the paxolin board, and marking through it with a sharply pointed tool. Drill holes for the tags of TC3, and for the wire ends of resistors, etc.
Insert the coils, resistors and capacitors, noting the positive end of C6 is as shown. Spread the wires to keep the components in position and turn the board over.
Items are now connected as in Fig. 3. Some 26 s.w.g. tinned copper wire, with 1 mm coloured sleeving will be most convenient. First join all the coil-can tags as shown, taking this circuit to the soldering tags MC. For easy checking, red sleeving can be put on all the chassis return, or positive, circuit wires.
Continue wiring, snipping off the unwanted ends of resistors, etc. Black sleeving can be put on the negative circuit connections, and yellow or some other colour on other wires. If each connection is marked with coloured pencil on the wiring diagram, as it is actually made; it extremely unlikely that any lead will be overlooked or wrongly wired.

The transistor leads can be identified by putting $\frac{1}{2} \mathrm{in}$. of black sleeving on the emitter wire, a similar length of yellow sleeving on the base, and red sleeving on the collector wire.

Fig. 3. Wirlng gulde for mixer-l.f. board.
Put the transistor leads in the correct holes, and draw them through until only the lengths of sleeving remain between the circuit board and transistor. Bend the leads over to the correct tags in Fig. 3, solder them, and cut off unwanted ends. A heat shunt is not essential if the iron is removed immediately the joint is made (a matter of perhaps two or three seconds). Fit the diode in the same way, with the polarity as shown.
The lead from the aerial will be soldered to Cl . Solder the wire to the junction of TC3 and C5, and identify it with coloured sleeving. Also solder a lead to pin 4 of the oscillator coil.
Take a black lead from R11, for the negative side of $\mathrm{Cl3}$. A connection is also soldered to diode D1 positive, and goes to VR1.
When the board is fitted to the chassis, the wires from pin 4 of the coil and diode D1 positive pass down through holes.

## AF amplifier

Components are fitted to an insulated board $3 \times 2 \frac{1}{8} \mathrm{in}$. as in Fig. 4. Note the polarity of C13, C14 and Cl 5 , and proceed as described for the i.f. board. The tag fixed with a bolt and lock nuts, providing the chassis return MC.
Provide colour coded leads from $\mathrm{Cl3}$ (negative supply to i.f. board) and C14 (audio circuit from R13 and volume control slider). Solder on a black flexible lead with a negative battery clip, and two leads from the secondary of T 2 , for the speaker.


* Fig. 4. Wiring guide for a.f. amplifier board.
components list

| Resistors: |  |  |  |
| :---: | :---: | :---: | :---: |
| R1 | 56k $\Omega$ | R12 | $1 \mathrm{k} \Omega$ |
| R2 | $10 \mathrm{k} \Omega$ | R13 | $2 \cdot 7 \mathrm{k} \Omega$ |
| R3 | $680 \Omega$ | R14 | 47k $\Omega$ |
| R4 | 3.9k $\Omega$ | R15 | 10k $\Omega$ |
| R5 | $56 \mathrm{k} \Omega$ | R16 | $560 \Omega$ |
| R6 | $680 \Omega$ | R17 | $1.5 \mathrm{M} \Omega$ |
| R7 | 1-2k $\Omega 5 \%$ | R18 | $470 \Omega$ |
| R8 | 3.9k $\Omega$ 5\% | R19 | 4.7k $\Omega$ 5\% |
| R9 | 8.2k $\Omega$ | R20 | 82, 5\% |
| R10 | $4.7 \mathrm{k} \Omega$ | R21 | $4.7 \Omega$ |
| R11 | $22 \mathrm{k} \Omega$ | R22 | $150 \Omega$ |

All $\frac{1}{4}$ W $10 \%$ except as noted
VR1, $5 \mathrm{k} \Omega$ potentiometer, $\log$, with switch, $\mathbf{S} 2$
Capacitors:

| apacitors: |  |  |  |
| :---: | :---: | :---: | :---: |
| C2 | $0.5 \mu \mathrm{~F}$ | C11 | 18pF 5\% |
| C3 | 180pF S.M. | C12 | $0.01 \mu \mathrm{~F}$ |
| C4 | $0.01 \mu \mathrm{~F}$ | C13 | $50 \mu \mathrm{~F} 12 \mathrm{VW}$ |
| C5 | 330pF S.M. | C14 | $4 \mu \mathrm{~F} 6 \mathrm{VW}$ |
| C6 | $8 \mu \mathrm{~F} 6 \mathrm{VW}$ | C15 | $50 \mu \mathrm{~F} 6 \mathrm{VW}$ |
| C7 | $0.1 \mu \mathrm{~F}$ | C16 | $0 \cdot 25 \mu \mathrm{~F}$ |
| C8 | $0 \cdot 1 \mu \mathrm{~F}$ | C17 | $50 \mu \mathrm{~F} 12 \mathrm{VW}$ |
| C9 | $0.1 \mu \mathrm{~F}$ |  |  |
| VC1/VC2 208-176pF (Jackson 00 Gang, threadedspindle) |  |  |  |
| TC1/2/3 60pF trimmer |  |  |  |
| Semi-conductors: |  |  |  |
| Tr1 | OC44 | Tr4 OC81D |  |
| Tr2 | OC45 | Tr5/6 OC81 | atched pair) |
| Tr3 | OC45 | D1 OA81 |  |

## Inductors:

Ferrite aerial MW/LW (Denco Ltd.)
L4 Oscillator coil (Weyrad P50/1AC)
IFT1 $465-470 \mathrm{kHz}$ (Weyrad P50/2CC)
IFT2 $465-470 \mathrm{kHz}$ (Weyrad P50/2CC)
IFT3 $465-470 \mathrm{kHz}$ (Weyrad P50/3CC)
Miscellaneous:
S1, Single pole, two way switch
T1, Driver transformer OC81 (Weyrad LFDT4)
T2, Output transformer $2 \times 0 \mathrm{C81}$ (Weyrad OPT1)
Chassis: $9 \frac{1}{2} \times 4 \times 1 \frac{1}{2} \mathrm{in}$. Speaker, $3 \frac{1}{2} \mathrm{in}, 3 \Omega$
Capacitor drive assembly (21-in. dia. drum, spindle, cord and spring) (Home Radio)
Flanged plate $5 \times 3 \mathrm{in}$. (Home Radio-CU145)
Paxolin sheet for circuit boards. Knobs etc.


Fig. 5: Wiring of panel controls. At right, arrangement of tuning drive assembly.

Several makes of driver and output transformer are available for the transistors used. In the case of the output transformer T2, connections to the centre-tapped primary, and secondary (for speaker) should generally be clear. With driver transformer T1, tag or lead positions, or colour coding, will depend on the manufacturer, so it should be checked against the maker's leaflet if necessary.

## Chassis and drive

Fig. 2 shows the top of the chassis, which includes the speaker and battery space, and which can be removed from the cabinet as a complete working unit. A hardboard panel, carrying the speaker and tuning scale, is secured to the chassis by S1, VR1, the cord drive spindle, and one or two bolts. An aperture is cut to match the speaker.
$\mathrm{VCl} / 2$ is bolted to a flanged plate 3in. high but short screws must be used. The drum is placed on the spindle with its flush side towards the capacitor and two ${ }^{3}{ }_{8} \mathrm{in}$. holes are punched so that the cord can run as shown in Fig. 5.

The variable capacitor has a threaded spindle, so that the pointer can be held with a 6BA bolt, small washers or spacers being put on to obtain a little clearance. The best method of assembly is probably by first preparing the chassis for SI, drive spindle, and VR1 and then fitting the ganged capacitor on its flanged plate to the chassis. Bolt on the supports for the ferrite rod.

Take the cord round the drum, through the chassis, give it one complete turn round the drive spindle, pass it back up through the chassis, and round the drum. Draw it to tension the spring, and tie, as in Fig. 5. Tighten the drum set-screw and check that the full 180 -degrees rotation is possible.

Temporarily remove the drive spindle nut and put on the prepared panel, holding it with this nut, S1, etc. Screw on the pointer, leaving a little clearance and then fix the loudspeaker.

Fig. 5 also shows connections to S1 and VR1 under the chassis. Solder a red flexible lead with a positive battery clip to the switch S2 of VRI.


## Assembly

It is best not to fit the i.f. and a.f. circuit boards until work on the chassis is completed. To fit the i.f. board, put $\frac{1}{2}$ in. bolts through the holes shown, and lock the nuts tightly against the soldering tags. Put further nuts on the bolts, which pass through matching holes in the chassis. These extra nuts are placed to give about $\frac{1}{4} \mathrm{in}$. clearance between the board and the chassis. A third nut is put on each bolt, and tightened under the chassis. If wiring is neat, no joints will be near the metal. However, a piece of card about $3 \times 4 \frac{1}{4} \mathrm{in}$. could be put under the board, before fixing it, to make sure no short circuit can arise.

When first testing the circuit, leave the collector end of R17, on the a.f. circuit board, disconnected. The board can be temporarily raised, with a lead clipped from chassis to MC tag. Temporarily connect R17 (Fig. 4), but if oscillation begins, or volume or distortion increases, switch off at once and connect R17 to the collector of the other output transistor. The correct connection is that which improves quality, and causes some drop in volume.

## Alignment

If a signal generator is available, proceed in the usual way for aligning a superhet.

To align without a generator, tune in any steady signal and then rotate the three i.f.t. cores, with a proper tool, for best volume. Repeat with a weak signal. Each core should have a definite tuning point giving best results.

Switch to m.w. and find a station with VCl/2 nearly fully open, and adjust TCl for best results. If TCl cannot be "tuned" to a peak giving best volume, modify the setting of TC3 until this can be done.

A signal is now tuned in with VC1/2 nearly fully closed, and L1 is moved along the rod, for best results.

Switch to l.w. and adjust trimmer TC2 and the position of L2 on the rod, for best results.

Repeat all adjustments (except those to the i.f.t.'s) until no further improvement is possible.

Actual frequency coverage can be modified by adjusting TC3, and the core of L4. After any such change, repeat alignment as described.

A scale is glued behind the pointer, and marked with station names, or frequencies or wavelengths determined by the signal generator.

Current drawn should be about $10-12 \mathrm{~mA}$ with no signal, rising to $20-45 \mathrm{~mA}$ with strong signals. A PP9 type battery is most suitable.

## Cabinet construction

Top, bottom and sides are made of $5 \times \frac{1}{4}$ in. sanded board which may be obtained from a handicraft supplier. The front is hardboard or 3-ply, cut so that it is a push fit. Drill clearance holes for the spindles, cut an aperture to match the speaker, and a window about $2 \frac{3}{4} \times 2$ in. for the tuning dial.

The parts are assembled with a quick-setting adhesive, and fine gauge panel pins, with the panel inset about $\frac{1}{4} \mathrm{in}$. The exterior is painted with clear cellulose.

A piece of fabric is cut and glued over the speaker aperture. A piece of Fablon-type material is glued over the remaining half of the panel and the dial window and spindle holes are cut through this with a sharp, pointed blade. A piece of glass or other transparent material is cemented inside over the tuning dial window.

A strip of coloured plastic or similar material about $\frac{3}{4} \mathrm{in}$. wide is cemented vertically to cover the joint between the panel finishing materials. Lengths of narrow beading are cemented round the panel edges. A plastic carrying handle is attached with screws or bolts.

The receiver is inserted from behind, and the spindles are sawn off to a suitable length for the control knobs. Brackets hold the chassis to the cabinet bottom.

# MARCH ISSUE, ON SALE FEBRUARY 5th 

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## PART 2



## HALLOR MOORSHEAD

Last month the basic operation was described and the circuits of the preamplifiers, record and monitor amplifiers given. The second part gives circuits of the oscillator and power supply and details the construction

## THE BIAS AND ERASE OSCILLATOR

Like the monitor amplifiers, the oscillator is based on a Mullard design. This has the advantage that no special inductors have to be used as the erase head inductance forms part of the network. A fair amount of power is needed for erasure and AD161-AD162 transistors are again used. As with the record heads the switching for either track or stereo has been left out to simplify the circuit but it basically involves either or both the heads. Astute readers will see that the bias frequency will thus be different on stereo or mono but it was found in practise that this difference was small. C21 and C22, C19 and C20 could of course be replaced by $0 \cdot 132 \mu \mathrm{~F}$ capacitors but as these are not available and even those near this value were hard to get, the two capacitors have been shown in parallel.

The value of R35 was found to be surprisingly critical; if it is too high the circuit will not operate and if it is too low the waveform takes on a very distorted form and this will cause all sorts of undesirable effects during recording. Three oscillators were built and the value shown worked well in all of them but if troubles are found with the oscillator this is the component to check and during the building of the Veroboard panel it may be as well to leave good long leads on this component.


As with the monitor amplifier panels two mounting holes are cut at the edges of the panel; no breaks are necessary in the copper strips. C27 is included to enable the erase to die away slowly once the supply voltage has been switched off, there is a danger otherwise of magnetising the tape heads.

C25 and C26 are the capacitors which couple a little of the a.c. signal to the record heads for the bias needed. 300 pF was found to be the best value in the prototype but this can bear experimenting with to obtain the best results.

## POWER SUPPLY

The power supply is conventional enough and is built on a separate chassis, the circuit is shown in Fig. 9.

It is well worthwhile fitting a fuse both in the mains supply and in the 22 V line, especially the latter. While checking the monitor amplifiers and oscillators individually, a 250 mA fuse can be used in the hope that this will protect the transistors in case of error. Once the amplifier is working correctly a 1A fuse can be substituted.

The bridge rectifier which is an LT120 type, available from Henry's Radio can, of course, be substituted by any with a working voltage of 30 V and a current handling capacity of 1A. Any transformer with a secondary in the order of 17 V or 18 V will do, although the one specified is widely available, easy to mount and gives a variety of other voltages apart from the one used.

## CONSTRUCTION

Almost as much thought was given to the constructional side as to the circuit design of the Tape Recorder. Several layouts and chassis designs were
※ NB. Erase heads are actually


Fig. 7: The circuit of the oscllator section.


The reverse of the main chassis. The oscillator section can be seen on the left with the monitor amplifier power transistors on the right.
considered and even tried but the one described here was finally chosen. Two chassis are used, one for the power supply and the other for the remaining circuit; both chassis are extremely simple but they allow for an uncrowded layout, can be wired easily, the component boards are accessible while the recorder is in operation, yet the appearance is attractive as can be seen from the photographs.

As mentioned two chassis are used; it is good practice to keep the power supply separate and as far away from the tape heads as possible as the magnetic fields surrounding the mains transformer can be picked up on the heads. To this end a small chassis holds the power supply at the back (or the top in case of vertical operation) of the cabinet. The extension loudspeaker sockets and the low level input/output sockets are also mounted on this.

The two chassis can be homemade and the overall dimensions are given, but arrangements have been made with H. L. Smith and Co. Ltd., 287 Edgware Road, London W. 2 to supply both units made from 16 s.w.g. aluminium for 12 s post paid. Heavy gauge metal is strongly recommended as the control panel is unsupported except along one long side.


Fig. 8: The layout of the oscillator Veroboard panel. No breaks are required in the copper strips.

## POWER SUPPLY CONSTRUCTION

The power supply should be built first as this will be needed for all later testing. Figure 10 shows the dimensions, drilling and basic wiring of this. The primary of the mains transformer should be wired for the local mains voltage and it is worthwhile


FIg. 9: The power supply c/rcuit.

far from this (outside the $16-18 \mathrm{~V}$ range) select two that give you the nearest to this (e.g. the 25 40 V may give 17 V rather than $15 \mathrm{~V})$.

These secondaries are wired directly to the a.c. tags on the bridge rectifier and the + and - connections are taken to the large smoothing capacitor C29. This component is mounted onto a short tag strip which also provides the chassis connection. The positive conection should be connected to the fuse.

Fig. 10: The construction of the power supply-the dimensions of the back panel are shown above.

## MAIN CHASSIS

The drilling of the main chassis is shown in Fig. 11 and the siting of the various components boards can be seen from the photographs.

The monitor amplifier, preamplifier and record amplifier and oscillator component boards are mounted about half an inch off the chassis by means of 6BA screws and spacing nuts.

If the boards have been built up first the holes previously
connecting the wires from the mains socket and the switched supply through a four way connecting block to provide a firm anchoring point for these wires.

A variety of secondaries are provided for on the specified transformer, 0-19-25-33-40-50V. Since we shall need 17 V a.c. this should be taken from the $33-50 \mathrm{~V}$ contacts. Before wiring these to the rectifier, however, it is worthwhile checking with a meter that you are getting 17 V since it seems that taps are not always accurate; if it is found that the voltage is


A rear view of the tape recorder. The plywood side panels and the speakers can be clearly seen with the power supply chassis propped up.
drilled for mounting can be used as a template for siting these screws exactly.

Once these positions have been marked the mica insulating washers for the power transistors can be used as a template, again for accurately siting the holes.


The P.W. Stereo Tape Recorder can be operated either vertically or horizontally as shown, but many decks are not suitable for vertical operation.


Fig. 11 : The basic drlling of the main chassis. See text for the siting of the mounting holes and the power transistors.

Once the drilling has been done the power transistors can be mounted, making sure that the bodies are insulated from the chassis. The collectors of three of the transistors are at chassis potential but do not be tempted to leave off the mica washers as this is exactly the sort of thing that causes instability.

The oscillator panel and the tags of the associated power transistor tags are mounted on the rear of the


Fig. 12 : The wiring of SWJ, the LEFT-STEREO-RIGHT swifch.
chassis; not only does this lead to a less crowded layout but it also tends to screen the oscillator from the rest of the circuit.

The various controls can be mounted after the power transistors. The specified V.U. Meters are held in position by a small amount of glue on the rear face. To save a lot of work later on cut the excess spindles from the controls as later on there is a lot of wiring that could be damaged by a slipping hacksaw.

Once the mounting screws have been fitted the main wiring can take place. It is very important to adopt a logical approach to this and the author used the rule LEFT-TOP-FRONT: this means that for wiring on all controls that the left hand channel wiring is always at the top and at the front where there is a choice. So the top track of the head is for the left hand channel, and for that same channel the record level control is at the front. It is amazingly easy to "cross" channels and this can lead to puzzling results (the author came across these several times).

There is one major earthing point and a common 22 V point. For these a tag strip should be fixed to the chassis just below the V.U. Meters. All earthing points and supply voltages should be wired to these. Assuming the power supply has been constructed first of all, it is best to build the monitor amplifiers


A view of the completed recorder, out of the cabinet with one of the speaker covers removed.

Once the various sections have been tested, the switch wiring can be tackled. No details are given in drawings as these will vary with the actual switch used.

The circuit of the switch for the Left-Stereo-Right switch is shown in Fig. 12 and again the wiring diagram of this has been omitted as the switch used will vary.

The DIN input sockets are wired directly to the input selector with the correcting components wired onto the sockets themselves.

A table of test voltages from the prototype is given for checking in case of difficulty.
next. These can be tested separately by injecting signals in at the volume controls. The input level and impedances are correct for a crystal or ceramic pickup and if one is used as the signal source the quality, level etc. can be properly tested. If an oscilloscope is available the signal across the loudspeaker can be monitored to see that there is no ultra-sonic oscillation. There is one great advantage in building stereo projects; each part is duplicated and if one section is not working properly one has something to compare it with.


Rear view of the completed recorder showing the access to the external connection sockets.

The emitter resistors for the output pair are wired directly from the circuit board to the appropriate tags on the transistors. Once the monitor amplifiers are working satisfactorily they can be used almost as signal tracers for the remaining circuit.

The preamplifier and record amplifier panels can next be mounted. For testing, the input side can be left unconnected, this being taken directly to the tape head, the output being fed directly to the input of the monitor amplifiers-once all this is working the majority of the circuit has been tested. One final test that will establish the working is to connect temporarily the output from the record amplifier to the input of the monitor amplifiers. This signal should sound very "toppy" and be at a much higher level than the proper output.

The oscillator can be checked by ensuring that there is some indication of a.c. volts across a tape head when the supply and heads are wired in. Most meters will be very inefficient at 50 kHz but most will give at least some indication that the thing is working.

CHECK VOLTAGES

|  | Collector | Base | Emitter |
| :---: | :---: | :---: | :---: |
| Tr1 | $1 \cdot 1$ | 0.3 | - |
| Tr2 | $2 \cdot 6$ | $1 \cdot 1$ | 0.5 |
| Tr3 | 8.5 | $0 \cdot 6$ | - |
| Tr4 | 9.0 | $1 \cdot 1$ | $0 \cdot 6$ |
| Tr5 | 9.0 | $0 \cdot 6$ | - |
| Tr6 | 22.0 | $9 \cdot 0$ | 9.0 |
| Tr7 | - | 9.0 | 9.0 |
| Tr8 | 22.0 | 11.0 | 11.0 |
| Tr9 | - | 11.0 | 11.0 |
| Voltages measured with a meter of $20 \mathrm{k} \Omega / \mathrm{V}$ under no signal conditions on record. |  |  |  |

It is vital that all wires carrying signals to the switches are screened and that all the earth points are taken to the common point. This is not really practical as far as the earthing of the DIN inputs are concerned nor is some of the screening of the wires to the Left-Stereo-Right switch, these exceptions are not important.

## CABINET CONSTRUCTION

The general construction of the cabinet can be seen from the photographs and the size and final finish are of course a matter of personal taste. The deck is mounted on two ${ }^{1}{ }_{2}$ in plywood pieces with a hole cut for the loudspeakers, the chassis is also screwed to this. This arrangement has considerable advantages from the constructional point of view but of course other designs can be used.

Once the Tape Recorder is completed make sure that the heads are aligned correctly before doing any serious recording and make sure that the erase heads and record/playback heads are in line.

The P.W. Stereo Tape Recorder has proved very satisfactory in operation and stood up well to continuous usage at the P.W. stand at the Audio Fair, despite the number of unwelcome dabbling fingers. When working the recorder will have a good quality specification and should be well worthwhile the trouble and expense.

## TEST-MASTER


#### Abstract

*CAPACITANCE SUBSTITUTION *gV SUPPLY *DIODE TESTER*RESSTANCE SUBSTITUTION *465kHz i.f. INJECTOR *MODULATED SIGNALS *AUDIO AMPLIFIER *1MHz SIGNAL INJECTOR *R.F SIGNALTRACER * CAPACITANCE BRIDGE * 1 kHz a.f. SIGNAL INJECTOR *OHMMETER *TRANSISTOR TESTER *RESISTANCE BRIDGE


IN recent years the author has done a considerable amount of dabbling in electronics and radio servicing and to this end has acquired a reasonable amount of test equipment. There is no doubt that a multimeter and an oscilloscope are by far the most useful pieces of test gear, but in addition there was a signal injector and tracer, a transistor tester, a small battery operated audio amplifier and a couple of cheap multimeters, one which was rarely used for anything else except to measure continuity on the ohms range and the other usually lashed into the supply circuit to monitor current consumption. In addition there was a box of components used for component substitution.
Most of the author's recent work has been with transistors and considerable use is made of S-Decs and T-Decs. These excellent products make possible reliable breadboarding and rapid development and unlike more traditional methods component leads are not messed up. A special tray was used for a long time, a sheet of plywood two foot square with an S-Dec glued to the centre, the outer rim of the tray being fitted with small cardboard boxes to hold various components. Unfortunately the more complicated projects meant that all sorts of test equipment were being used in conjunction and the old time "rats nest" was being replaced by another.

To overcome this the author decided to build a small "laboratory" into a conveniently sized case which would replace the majority of the bits and pieces used. It was also decided that facilities should be included to aid servicing and in practice the final result has made this very much easier.

## CHOICE OF FACILITIES

There are some facilities that are essential to a unit of this type; both the small audio amplifier and the signal injector were constantly in use and it was decided that these were obvious candidates for inclusion. Resistance and capacitance substitution were also thought to be very useful as having these on
switches would allow immediate selection of a wide variety of components.

An ohmmeter was also thought very usefulespecially for rapid checking of continuity and since a meter was necessary for this a few extra components were incorporated to allow the internal battery to be available for use with a 100 mA meter in series to monitor the current consumption. The meter is also used for testing transistors, both p-n-p and $n-p-n$ for leakage and gain.

As standard value capacitors, an audio generator and an audio amplifier were already to be included, it was thought worthwhile to include a capacitance bridge. The unit will also work as a resistance bridge but this facility is less useful.

The signal inject facility eventually decided upon was to have an audio note available of about 1 kHz and to have an r.f. signal of about 1 MHz as well as an i.f. signal of 465 kHz , the last two being capable of being modulated by the 1 kHz audio note.

The audio amplifier would have to be fairly sensitive and in fact will produce a useful output from a 0.5 mV signal. A simple switch was included at the input to include a detector for monitoring r.f. or i.f. signals.

It is of course a fairly simple matter to build the various items mentioned above, but the virtue of the unit described is that by the turning of one master switch, all the facilities mentioned above are available between two probes. The alternative would be to have a series of terminals on the front which would have to be changed with each function and this would have destroyed one of the major advantages.

The block diagram, which gives the basic circuitry, is shown in Fig. 1 and the various operations are as follows:

## SUBSTITUTION RANGE

The negative terminal of the probes is connected at all times, directly to the aluminium chassis which


Fig. 1 : The basic circuit of the Test-Master showing all the switching. The circuits of the three modules are shown overleaf.
acts as the common line, the positive terminal being connected to one section of SW1. When this switch, which is the main selector, is set to "substitution," the positive probe goes directly to the wiper of a one-pole, twelve-way switch. Eleven of these positions are wired to a variety of capacitors whose other ends go to the chassis. The actual choice of capacitors is a matter of personal preference but in the prototype the lowest value is 100 pF followed by $1,000 \mathrm{pF}$ and there on up values increasing by a factor of three in each stage to $100 \mu \mathrm{~F}$. Values below $1,000 \mathrm{pF}$ are generally used for r.f. work and become quite critical, in addition there is a danger of radiating signals because one end is connected to the chassis. Also stray capacity would vary the lower values. Values of $1 \mu \mathrm{~F}$ and above are electrolytic types and their working voltage is again a matter of personal preference. Ideally they would all be high but this would mean the use of very large components. The prototype uses 25 V or higher components which enable them to be used with most transistor equipment.

The twelfth position on the capacitance switch connects to the resistance substitution switch, which, like the previous one, is a one-pole twelve-way type. Values here run from $3 \Omega$ to $3 \cdot 3 \mathrm{M} \Omega$ and were chosen so that differences between one value and the next increase by a factor of 3 . Again the wattages chosen depend on the likely usage; the prototype used ${ }_{1}{ }_{8}$ watt values for all except $3 \Omega$ which was included for loudspeaker loads and the $3 \cdot 3 \mathrm{M} \Omega$ which is not available in $1_{8}$ watt, a $1_{4}$ watt type was used instead. Like the capacitors the other connection of the resistors are joined together and go to the chassis.

## AUDIO AMPLIFIER

In position 2 of SW1 the positive probe is connected to SW4 which is the normal/detect switch for the audio amplifier. SW1b applies 9 V to the amplifier.

SW4 in the "Detect" position applies the signal to the top of VR1 via an OA91 diode, the other part of the switch connecting a $1,000 \mathrm{pF}$ to earth to smooth the signal. In the "Normal" position the diode is bypassed and C2 disconnected. C12 only serves to d.c. block the input. A highish value potentiometer is chosen for VR1 as this allows high impedance signals to be amplified with the slider near the middle.

The circuit of the amplifier is shown in Fig. 2 and it will be seen that it is very simple. High quality is not all that important but nevertheless some care
was taken in the design to ensure high gain and stability. To improve stability a small capacitor decouples the very high frequencies from the base of the output transistor. A high impedance speaker ( $35 \Omega$ $80 \Omega$ ) is used as this avoids the necessity for an output transformer. The circuit used is fairly conventional and has the virtue that the application of bias to the first stage from the emitter of the second stage d.c. stabilises the circuit and makes allowances for the wide divergence of parameters associated with cheap transistors. 2N2926G transistors were used but many similar silicon types will do just as well with no alteration of component values.

## SIGNAL INJECTOR

Although one section on the main selector switch is marked "Inject" this only leads to a subsidiary switch, SW2 which selects the output from one of two oscillators, either the r.f. or the a.f. A second section of this switch alters the frequency of the r.f. oscillator from about 1 MHz to 465 kHz . A further subsidiary switch, SW5 provides modulation for the r.f. and i.f.,signals.

The audio oscillator is a simple Hartley type, the inductor being a small tapped output transformer, normally used in push-pull output stages. The circuit is shown in Fig. 3. The centre tap goes to the


Fig. 3: The audio oscillator. This produces a 1 kHz note for signal injecting and for the $R-C$ bridge.


Fig. 2: The audio amplifier circuit module.
positive supply line. The value of the capacitor across the total primary determines the frequency and can be varied to suit. In the prototype the transformer used with the specified capacitor gave a frequency of about 1 kHz and this is about right. R21 provides the base bias and C25 the feedback.

In the normal inject mode the output is taken via C26 from the collector. The secondary of the transformer is used for the bridge circuit only, this providing a balanced output. The shape of the audio note hardly
matters in the inject mode and a simple multivibrator or relaxation oscillator could have been used but a decent sine wave is far better for use with a bridge circuit.

The r.f./i.f. section is almost identical, a Hartley type circuit being used again but instead of an output transformer an i.f. transformer is used. The circuit is shown in Fig. 4. A 465 kHz transformer could have been used here and this would make the i.f. injector very simple but to obtain a fundamental fre-


Fig. 4 : The r.f./i.f. oscillator module.
quency in the r.f. spectrum it would necessitate removing the small tuning capacitor inside the i.f. can and arranging for this to be switched.

This was originally tried but it was a messy business and it was instead decided to use a 1.6 MHz i.f. and to put capacitors in parallel with this to lower the frequency to 1 MHz and 465 kHz . However clumsy this solution sounds it does in fact make construction far easier. To be able to make this work the specified i.f. transformer must of course be used.

With the internal capacitor and the two external capacitors necessary to produce the 465 kHz oscillation it will of course be necessary to trim the core of the coil to obtain the exact i.f. frequency required. Obviously this means that the 1 MHz signal will not be exact but this doesn't really matter as it is only intended for general r.f. use. It is, of course, possible and sometimes desirable to set the i.f. frequency to 455 kHz , the standard Japanese/U.S./Hong Kong transistor radio i.f., but this is a matter of choice.

## TRANSISTOR TESTER

Unlike the other functions of the unit, the transistor tester section is operative at all times-this is quite acceptable as no current is drawn from the batteries if no transistor is fitted. The basic circuit is shown in Fig. 5 but the complete circuit can be easily traced from Fig. 1.


Fig. 5: The basic transistor tester circuit.


The photograph shows the wirlng used for the substitution range switches.

Referring to Fig. 5 it will be seen that with the switch open the battery is applied in the correct polarity across the transistor. With this set-up the meter will register the leakage passed by the transistor. As the ohmmeter resistors R14 and VR2 are connected in series the readings will be logarithmic and they will also protect the meter when a shortcircuit transistor is inserted.
As a general guide one can say that the first 'third' of the scale registers acceptable leakage, the second 'third' inferior, but usable, devices and the final 'third' duds. However certain transistors, especially power output germanium types will show high leakage while still being perfectly O.K. and so the readings are subjective and one will have to have tested quite a few good types before the readings tell you anything. This test is quite crude but it is rare that it will not tell you if a transistor is usable.

When the switch is made a $100 \mathrm{k} \Omega$ resistor is connected between the collector and the base thus providing bias and the transistor will begin to conduct. The reading shown on leakage will rise appreciably on a good device and the rise will give an indication of the gain. Once again readings and their interpretation are a matter of judgement.

Low leakage (especially silicon types) transistors will show little or no reading on leakage and this will not mean that the device is open-circuit.

SW3 is a 3-way, 4-pole switch wired as in Fig. 6 and this serves to change the battery supply to the


Fig. 6: The wiring of the transistor test switch.
transistor test socket and also to apply base bias in two of the positions.

Diodes can be tested quite easily using the same basic circuit. When connected one way around be-
tween the emitter and collector junctions the leakage shown should be very low-no indication at all is quite common. When the device is reversed a pretty high current will be registered on the meter-this of course applies to a good device only.

## BATTERY SUPPLY

In the battery supply position of SW1, the positive connection is coupled through the meter with a section on SW1b shunting the meter so that it reads 100 mA f.s.d. The value of R 13 should be one-hundredth of the internal resistance of the meter; in the case of the one specified this is $109 \Omega$. The author


Fig. 7: The method of obtaining the correct shunt resistor to make the meter read 100 mA f.s.d.
was lucky enough in finding a resistor that was exactly right but it is possible to derive the value as follows.

Initially choose a resistor that is slightly over $1 \Omega$ and shunt this across the meter and then couple this in series with a $100 \Omega$ potentiometer. The additional components are shown dotted in Fig. 7. Set the pot. so that 100 mA is shown passing in the circuit, this being read off the testmeter which should be on a suitable range. The meter that we are trying to shunt should, in this position, have its needle pressing hard against the back stop. When this arrangement is achieved start putting low value resistors ( $10 \Omega$ to $100 \Omega$ ) in parallel with R13 until one is found that shows f.s.d.

It doesn't matter greatly about the accuracy of the supply current reading and many readers will be content with using a $1 \Omega 5 \%$ resistor-this will give an accuracy $\pm 7 \%$ which, although not good enough for normal measuring standards, should generally be good enough to tell us if something is wrong with a piece of equipment.

It is good practice to have a meter in the battery supply to transistor equipment when working on it as it will show up shorts and transistors running away in time to stop any damage and is a very much more useful facility than at first appears.

## OHMMETER

The circuit of the ohmmeter is shown in Fig. 8. This


Fig. 8 : The ohmmeter circuitry.
is a standard circuit with VR2 acting as the "Zero Ohms" control. As a 9 V battery is used this is a fairly sensitive ohmmeter and the centre scale deflection, which is the only proper way of giving it's sensitivity, is $18 \mathrm{k} \Omega$. This is somewhat overcome by the small scale of the meter itself but it should nevertheless be very useful. It will have to be calibrated using known value, close tolerance resistors but if it is only to be used as a continuity tester there will be no need.


An internal rear view of the Test-Master. The battery fits into the space on the right below the meter.

## BRIDGE CIRCUIT

Although bridge circuits usually require an article to themselves we have already got nearly everything necessary to incorporate one. The a.f. oscillator provides a balanced output from the secondary of the transformer used, the amplifier can be used as the detector and the capacitors used for substitution .can be used as the standards against which to balance the unknown value. All that is necessary to add is a $5 \mathrm{k} \Omega$ potentiometer.

At a particular setting of the potentiometer the bridge will balance and there will be only a low output from the loudspeaker and the setting of that pot. will give a multiplication or dividing factor with which to calculate the value. The actual arrangement used here is especially useful as the value of the capacitors go up by a factor of about three each time and so the calculating factor will never be more than this.
By changing the capacitance switch to the resistance substitution range it becomes a resistance bridge and this can be useful for measuring low value resistors and possibly high value ones as well (though the simple design means that breakthrough can be troublesome here).
The switching to achieve this arrangement can be


Fig. 9: The bridge circuit for measuring the values of unknown capacitors.
traced from Fig. 1, the basic circuit is shown in Fig. 9 and this will be recognised as a conventional bridge circuit.

## CONSTRUCTION NOTES

The actual construction of the Test-Master should present no real problems, the only ones are likely to be getting the switches wired correctly.

A large aluminium chassis, 12 x $5 \times 2^{1} 2^{\prime \prime}$ is used on its side so that what is normally the top becomes the control face-quite a lot of controls are involved as can be seen from the photographs but little depth is needed. A smaller cabinet could be used-there is little crowding of components-but this will make wiring difficult.

There are three "active" parts to the Test-Master; the audio amplifier, the a.f. signal source and the r.f./i.f. signal generator. These are all built on small Veroboard panels and mounted on the bottom and side of the chassis.

The component layout on the panels is shown in Figs. 10, 11 and 12 together with the connections to the adjacent switches etc.

The positioning of the meter, the controls, switches and loudspeaker are shown in Fig. 13; the sides are shown in the same plane as the front. Exact positioning will depend on the type and size of the components and so dimensions are not given but some thought was given


Fig. 11 : The component layout of the audio amplifier.


Fig. 12 : The r.f./i.f. injector module.


Fig. 13: Approximate siting of the controls and modules inside the chassis.


Fig. 10: The layout of the audio oscillator module.
to the arrangement of the various sections and the layout shown does give relatively short wires to the various sections.

The battery, a PP7, is mounted beneath the meter. The panels themselves can be laid on a sheet of card


An internal view of the completed Test-Master.
(to stop the bottom connections shorting to chassis) with the screws through the mounting holes clamping them to the chassis.

The socket associated with the transistor tester is a normal transistor mounting socket; a slightly

## $\star$ components list

| Resistors : |  |  |  |
| :---: | :---: | :---: | :---: |
| R1 | $3 \cdot 3 \Omega 2 \mathrm{~W}$ | R12 | 3.3M $\Omega$ |
| R2 | $33 \Omega$ | R13 | $1 \Omega$-see te |
| R3 | $100 \Omega$ | R14 | 3.9k $\Omega$ |
| R4 | $330 \Omega$ | R15 | 100k $\Omega$ |
| R5 | $1 \mathrm{k} \Omega$ | R16 | $10 \mathrm{k} \Omega$ |
| R6 | $3 \cdot 3 \mathrm{k} \Omega$ | R17 | $1 \mathrm{M} \Omega$ |
| R7 | 10k $\Omega$ | R18 | 56, |
| R8 | $33 \mathrm{k} \Omega$ | R19 | $680 \mathrm{k} \Omega$ |
| R9 | $100 \mathrm{k} \Omega$ | R20 | $2.7 \mathrm{k} \Omega$ |
| R10 | $330 \mathrm{k} \Omega$ | R21 | $270 \Omega$ |
| R11 | $1 \mathrm{M} \Omega$ |  |  |
| All resistors 5\%, $\frac{1}{8}$ watt except R1. |  |  |  |
| VR1 VR2 | $250 \mathrm{k} \Omega$ log. $5 k \Omega$ lin. | VR3 | $5 \mathrm{k} \Omega$ |
| Capacitors : |  |  |  |
| C1 | 100pF | C14 | $0.1 \mu \mathrm{~F}$ |
| C2 | 1000pF | C15 | $10 \mu \mathrm{~F} 12 \mathrm{~V}$ |
| C3 | 3300pF | C16 | 1000pF |
| C4 | $0.01 \mu \mathrm{~F}$ | C17 | $10 \mu \mathrm{~F} 12 \mathrm{~V}$ |
| C5 | $0.03 \mu \mathrm{~F}$ | C18 | 360 pF |
| C6 | $0.1 \mu \mathrm{~F}$ | C19 | 1800 pF |
| C7 | $0.33 \mu \mathrm{~F}$ | C20 | 33pF |
| C8 | $1 \mu \mathrm{~F} 25 \mathrm{~V}$ | C21 | 1000pF |
| C9 | $3 \cdot 2 \mu \mathrm{~F} 25 \mathrm{~V}$ | C22 | $80 \mu \mathrm{~F} 12 \mathrm{~V}$ |
| C10 | $32 \mu \mathrm{~F} 25 \mathrm{~V}$ | C23 | $10 \mu \mathrm{~F} 12 \mathrm{~V}$ |
| C11 | $100 \mu \mathrm{~F} 25 \mathrm{~V}$ | C24 | $0 \cdot 1 \mu \mathrm{~F}$ |
| C12 | $0.1 \mu \mathrm{~F}$ | C25 | 2200pF |
| C13 | 1000 pF |  |  |

Switches :
SW1 3-pole, 6-way rotary
SW2 2-pole, 3-way rotary
SW3 3-pole, 4-way rotary
SW4 2-pole, 2-way toggle
SW5 On-off toggle or same as SW4
SW6 1-pole, 12 -way rotary
SW7 1-pole, 12-way rotary
Semi-conductors :

| Tr1 | 2N2926G | Tr4 | 2N2926G |
| :--- | :--- | :--- | :--- |
| Tr2 | 2N2926G | D1 OA91 or similar |  |
| Tr3 | 2N2926G |  |  |

## Miscellaneous :

Meter- 1 mA d.c., Henelec ' 38 Series'.
Loudspeaker-miniature (2-21 i in. diameter), $35 \Omega-80 \Omega$ impedance.
T1-tapped transistor audio output transformer (Henry's Radio type JOT).
I.F.T.1-1.6MHz i.f. transformer, Denco type IFT. 16/1.6.
Veroboard-3 off panels $11 \times 9$ holes, 0.15 in . hole matrix.
Aluminium chassis- $12 \times 5 \times 2 \frac{1}{2} \mathrm{in}$. - see text.
Transistor mounting socket.
Output/Input sockets.
undersized hole will enable it to be pressed firmly home.

For the substitution ranges the components can be mounted as shown and soldered in place; with the resistors which are all the same size this is easy but the bulkier capacitors will have to be arranged as well as possible.

Two-way, two-pole toggle switches were preferred
over miniature slide switches as they are far easier to mount. The loudspeaker for the amplifier is best glued into position behind a small expanded metal grill.

For readers who have difficulty making things work first time (and aren't most of us in that category) it is fairly simple to fix and test one facility at a time; for instance wire up the main switch first with flying leads and then only try the capacitance substitution range followed by resistance substitution and so on ending up with the bridge circuit.

As far as wiring is concerned, screened wire may be better for the outputs of the signal injectors and the input to the audio amplifier but the author was prepared to accept a certain amount of breakthrough and so normal wires were used throughout.

No facility is incorporated for attenuating the level of the injected signals and a further control could be inserted for this purpose but it was found easier to have an extra probe with à $1 \mathrm{M} \Omega$ resistor incorporated as this can also be used for tracing high level signals.

Once completed the Test-Master should prove to be a really useful addition to any test bench and will enable a lot of odd pieces of test gear to be relegated to the shelves for use on rare occasions only.

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JUDGING by the appearance of the average radio enthusiasts den, shack, workshop or what-haveyou the percentage of home-made equipment that is properly finished off is quite minimal.
The author is the first to admit that he was the world's worst offender in this respect, having over the years built just about everything from an absorption wavemeter to an all-band SSB transceiver. The gear always worked, eventually, but having spent so much time on the building of it the author was usually only too happy just to sit back and play with the knobs for a while!

The thought of taking off that panel with its myriad knobs, dials, switches, etc., was anathema and anyway there was always that new project that had to be started!

It is a good idea to relax once a project has been completed but it is also necessary to be quite determined, in the near future, to put that finishing touch to it so that the uninitiated may remark on the excellence of one's handiwork whether it actually works or not!

Some enthusiasts have even reported receiving actual words of encouragement from their dear ones on emerging from the shack with a new bit of gear that was actually in a cabinet and which had a panel that was not dented or scratched or a mass of unused holes, but such reports should be treated with some reserve!

But seriously, a great deal of satisfaction is to be had by spending a little time and effort on putting the finishing touches to a job and it need not be an expensive or lengthy procedure.

The legendary rift between the enthusiast and his family does, one can be sure, spring from the good old days when to get a good crackle finish on a panel it was necessary to bake the panel in the kitchen oven for a couple of hours! The fact that the roast joint tasted of paint for the next fortnight was quite immaterial!
Fortunately today paints are available that dry hard in an hour or so and which can be sprayed from a can. A variety of colours gives a wide choice to satisfy individual tastes.

## FINISHES

Among the many possibilities should be mentioned the natural metallic finish, painting by brush or spray, the covering of the panel with a plastic film and the use of a dummy panel attached to the front of the main panel.
Finishing should also take into account the lettering and numbering of all the controls, switches, etc. on the panel as one would expect to find on commercial equipment. This again is not difficult to achieve and one does not have to be an artist to stick a few transfers on to a panel.

## PREPARATION

Whatever finish it is decided to adopt there are certain preliminary steps that are essential to ensure good results. If the design of the equipment is being copied from a magazine the chances are that the panel layout will be given and that the author has made some effort to ensure that the layout of the controls is a reasonably symmetrical one.
If the constructor has to make his own layout he should spend some time on working out a neat arrangement for the controls consistent with the electrical requirements of the circuitry. In marking out the final positions of holes, cutouts, etc., use a rule and pencil and do not "guesstimate" as the eye can often be quite unreliable as far as measurements are concerned.
After the holes have been made ensure that all burrs and rough edges are removed from both sides of the panel.
Next, lay the panel on a sheet of thin cardboard or stout paper and with a sharp pencil make an outline drawing of the panel and all holes and cutouts. Be careful about using a biro pen for this purpose as the tip is usually quite blunt and the resulting drawing will not be at all accurate in size.
This drawing will be found very useful for future reference especially if it is decided to fit a dummy panel to the equipment as will be described later.

## NATURAL FINISHES

An aluminium panel may be given quite an attractive finish by laying it on a sheet of newspaper and rubbing the panel down on one side with a piece of wood around which a piece of fine sandpaper has been wrapped. A fine oil can be used as a lubricant during the rubbing down but great care must be taken to ensure that the oil is removed completely when the panel is finished. A circular rubbing motion will produce the best effect.

To complete this particular finish the panel should be given a coat of transparent varnish with a soft brush or from an aerosol spray.

It is important to read the instructions on these spray cans in order to get a smooth clean result. Often the spray is held too near the object or is not moved about fast enough and blobs and runs appear in the finish.

## PAINTING

The number and types of paint available today are legion but attention should be paid to those paints that are available in aerosol spray cans. A large range of colours can be found in most motor accessory shops as well as in the usual paint and decorators and DIY shops. The sprays are cheap, clean,
fast and just about infallible if the instructions are followed.

Again, the panel should be laid on a sheet of newspaper and carefully cleaned of all dirt or grease with a white spirit or similar solvent. Spray the panel according to the instructions and remember that two thin coats are better than one thick coat. So do not try to put on too much paint in one go but apply a thin coat and allow to dry properly before applying the finishing coat.

It is a good idea to do the whole job outdoors if the weather permits as otherwise the fumes from the paint can be very objectionable and even dangerous indoors.

Give the paint ample time to dry before starting to reassemble the panel and above all don't be tempted to prod the panel with a fingertip 'just to see if it is dry' . . . it won't be! The drying process can be speeded up by directing the heat from an electric fire on to the panel for a short time.

Air-drying crackle finish paint such as Yucan is now obtainable in aerosol spray cans in various colours and this provides an excellent finish that will hide blemishes on a panel that might not be so easily covered by a flat paint.

## PLASTIC FILM FINISH

This method of finishing a panel is one of the simplest, cheapest and cleanest ways and the time spent in removing and replacing the panel on the equipment is well justified.

The plastic film is available from Letraset dealers and art shops under the name of "Letrafilm" in no less than 119 colours in matt finish and 20 colours in a gloss finish. Sheets $10^{\prime \prime} \times 7^{1}{ }^{\prime \prime}$ " are 4 s 3 d each and $20^{\prime \prime} \times 15^{\prime \prime}$ at 8 s 6 d each.

Remove the panel from the equipment checking that there are no burrs or rough edges. These are easily caused by the action of lock washers especially on soft aluminium panels.

Lay the panel over, one corner of the sheet of Letrafilm, moving the panel around to get the most economical position. Offcuts of the material can come in very handy. Mark round the outside of the panel with a pencil leaving a border of about $\mathrm{I}_{4} \mathrm{in}$. and then cut out the piece required with scissors.

Remove the backing paper from the film and drop the centre of the film on to the centre of the panel. Drop one end and smooth it down then dropping the other end and smoothing that down also. With this technique there will be a minimum of air bubbles underneath the film. Finally put a sheet of clean paper over the whole panel and with the ball of the hand smooth the film down leaving a perfectly flat and bubble-free surface.

The film may be pushed through small holes using a biro pen. On larger holes cut the film in the hole radially and fold back behind the panel. The excess material at the panel edges may now be folded over the edges, if necessary snipping the film at the corners.

The whole operation takes far longer to describe than to perform!

## DUMMY PANELS

The advantage of a dummy panel is that it is not necessary to take off the main panel but only to
remove the locking nuts and fixing screws from the components mounted on the panel. We shall now need the tracing of the front panel that it was recommended should be made once the main panel is completed.

The dummy panel may be plàstic sheet of the appropriate colour and about $1 / 32$ in. thick. Place the plastic sheet underneath the tracing and mark through the outline of the panel and all the holes and cutouts, then cutting them out with scissors or a sharp knife such as is used by modelmakers.

Panel screws should be countersunk where possible so that they will be out of sight when the dummy panel is fitted.

If the main controls are reasonably spread out over the area of the panel the holes nearer the centre can be clearing holes around locknuts, etc., thus simplifying the work even further.

## PANEL LETTERING

The final touch that cannot be omitted on any score, least of all of expense! Sets of panel transfers can be obtained from Data Publications, 57 Maida Vale, London, W9 at 5 s . per set. These cover receiving, transmitting and audio equipment requirements.

The transfers are easily applied by following the instructions supplied. But one tip, if there are to be several transfers in a line rule a faint pencil line on the panel in order to assist the placing of the transfers on to the panel.

It is better if the transfers are put on before the panel is replaced on the equipment and, of course, after the panel has been painted or otherwise finished off, but don't forget to take into consideration the diameter of scales, knobs, etc. before finally placing the transfers.

Another point worthy of some thought is that of placing the transfers above or below the controls. If the equipment is to be low down the transfers should be over the controls so that they can be read from above. If the equipment is, say, up on a shelf then the transfers ought to be below the controls.

## GENERAL NOTES

Whether a panel is to be properly finished off or not there is a right and wrong way of mounting components on a panel. Subject to space permitting, potentiometers and switches etc., should have two lock nuts fitted. This allows the minimum of the threaded bushing to project at the front of the panel enabling knobs and dials to lay as close to the panel as possible. It also removes strain from the body of the component itself, and finally, it is good engineering practice.

Do not be shy about cutting off any excessive length of spindle. This is best done by holding the end of the spindle in a vice or pair of pliers and cutting off the excess spindle with a fine-toothed small hacksaw afterwards removing any burrs with a file. Do not make the mistake of holding the component itself in the vice and possibly damaging it irreparably.

When refitting potentiometers and rotary switches to the panel make sure that the arc of rotation is symmetrical so that the indicating line or spot on the knob is vertical when in the centre of its travel. A small point perhaps but it does add to the final finish of the equipment.
-continued on page 854

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$350-0-35080 \mathrm{~mA} .6 .3$ v. 3.5 a. 6.3 v. 1 a 5 . $350-0-35080 \mathrm{~mA} .6 .3$ v. 3.5 a. 6.3 v. 1 a, or 5 v. 2 a. MINLATURE $200 \mathrm{v} .20 \mathrm{~mA}, 6.3 \mathrm{v} .1 \mathrm{a}, 2 \frac{1}{4} \times 2 \times 1 \frac{5}{5} \mathrm{in}$. MNLAT
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## SHORTWAVE TRANSISTOR RADIO

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"READ PEOPLE'S MINDS"--TEST THEIR NERVES —BEAT THEM AT CARDS-TEST THEIR MEN. TAL ABILITY, ETC. WITH THIS ASTONISHING ELECTRONIC BRAIN-BOX
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Fortunately, it is possible to construct a useful keyer around a single transistor plus a handful of small components and if care is taken excellent results can be obtained; in all fairness, however, it must be pointed out that it is the more refined type of keyer that does the better job for the serious operator.

## Circuit

The circuit of a keyer which has been in regular use for over a year and which is complete with its own audio monitor is shown in Fig. 1. The basic switching circuit is by no means unique and to it is added the essential variable speed facility plus provision for adjusting the dot-dash length; a degree of dot-dot space control is also incorporated.
Transistors $\operatorname{Tr} 1$ and $\operatorname{Tr} 2$ plus the associated items are the elements of the audio monitor, whilst the section around Tr 3 takes care of the keying.
Adequate switching is also provided and when ganged switches Sla and Slb are at the position

shown in Fig. 1 the keyer is off. With switch Sl in position " 2 " the battery negative is applied to the paddle control blade via one section of the relay but transistor $\operatorname{Tr} 3$ remains inoperative since its base is open circuit.

When the paddle control is held against the "dot" stop, however, the transistor base goes heavily negative. Tr 3 conducts hard and capacitors C 3 and C 4 are charged. The relay operates and its contacts open which immediately cuts off the base supply potential. Capacitors C3/C4 immediately start to "discharge via R6 and VR1 but the relay is "held on" for a period of time dependent upon the setting of VR1. When C3/C4 have discharged sufficiently the relay drops out and its contacts change over once more. If the paddle is still in contact with the dot stop the operation is repeated and continues until the control is returned to the central resting position. Gate diode Dl does not conduct during the dot formation.

When the control paddle is held against the dash stop, gate diode Dl conducts as its cathode end goes negative with respect to its anode. With the necessary negative potential applied to the transistor base the relay again operates. The base circuit discharge period is longer now, however, for additional capacitance due to C2 (via D1) is introduced and this ensures that the relay cannot drop out as quickly as it did during the dot forming operation; a dash is thus produced.
Capacitor C3 and control VR2 are included to make the dot-dash ratio adjustable over a small range. The repetition frequency is decided mainly by the setting of VR1, however, and this is a panelcontrolled item marked "Speed". A degree of controllable mark/space ratio is afforded by VR4, being so set that the keyer just operates.

Fig. 1. Complete circuit of the solid-state keyer.

## Monitoring

Whenever the control paddle is at its central rest position current flow is negligible even when S1 is at position " 2 " and the collector current of $\operatorname{Tr} 3$ can be measured in micro-amps. The audio oscillator around $\operatorname{Tr} 1$ and $\operatorname{Tr} 2$ is also inoperative provided the relay is connected with its contacts as shown when in the rest position.

Immediately the keyer is operated, however, a negative potential appears at the junction of R4/LS and this energises the audio oscillator. The audio oscillator proper is that section around Trl, transformer T1 providing the required degree of feedback essential for oscillation from collector to base. A crude tone control VR3 in the emitter circuit allows the note obtained to be adjusted within limits to suit the user's taste.

Transistor $\operatorname{Tr} 2$ merely acts as a simple amplifier feeding the loudspeaker LS which may be any medium impedance insert type. A balanced armature type works well but sub-miniature speaker units or discarded headphone elements may be tried. Transformer TI is a sub-miniature type which measures but $\frac{1}{2} \times \frac{3}{8} \times \frac{1}{2}$ in and has a ratio of $5: 1$ with a primary inductance of 12 H .

Since the audio generator energising potential is tied to the relay operation the monitor note is heard in sympathy with the keying. It may also be noted with interest that the monitor note can be obtained by keying irrespective of whether the transmitter is connected or not and this means that the unit can be employed for Morse code practice if required!

Note that a further position is used on switch Sl which when switched to " 3 " leaves the keyer dead but short-circuits the output and gives "Key down" for netting or tune-up.

## Construction

A standard Radiospares 18 -way tag board is a convenient base on which to construct the keyer. For controls VR2 and VR4 miniature pre-set items are suitable, but for the speed control VR1 a long spindled item must be used to enable the control knob to be brought clear of the final housing. The keyer outlet socket can conveniently be made a coaxial type; such a socket is useful in that it allows various transmitters to be connected not all of which have standard jack sockets-the Codar AT5 for example.

A "U" bracket for the paddle contacts is easily fabricated on the lines shown in Fig. 2, the saw cut being made last. Adjustable dot and dash stop terminal screws can be inserted and tightened in the holes in the uprights. The tag board may be held securely to a metal base using a few lin spacers cut from ball-point pens.

The paddle control blade is made from thin brass strip and the ideal material is to be found in the composing departments of printers where a piece of 25 "ems" ( $4 \frac{1}{\frac{1}{8}}$ in) $\times 1 \frac{1}{2}$-point "brass rule" will if cut

## Resistors:

| R1 | $18 \mathrm{k} \Omega$ | R4 | 2.2k $\Omega$ | R7 | 10k $\Omega$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R2 | $4.7 \mathrm{k} \Omega$ | R5 | $470 \Omega$ | R8 | $1 \mathrm{k} \Omega \frac{1}{2} \mathrm{~W}$ |
| R3 | 270k $\Omega$ | R6 | $1 \mathrm{k} \Omega$ | R9 | $390 \Omega$ |
| VR1 | $10 \mathrm{k} \Omega$ |  | VR4 | $500 \Omega \mathrm{p}$ | re-set |
| VR2 | $5 \mathrm{k} \Omega$ | -set | VR5 | $1 \mathrm{k} \Omega$ po | wirewound |

VR3 $10 \mathrm{k} \Omega$ pot
Fixed resistors $\frac{1}{4}$ W $10 \%$ except R8
Capacitors:

| C1 5000 pF ceramic | C 4 | $10 \mu \mathrm{~F} 25 \mathrm{~V}$ |
| :--- | :--- | :--- |
| C 2 | $25 \mu \mathrm{~F} 25 \mathrm{~V}$ | C |
| $100 \mu \mathrm{~F} 50 \mathrm{~V}$ |  |  |
| $\mathrm{C} 310 \mu \mathrm{~F} 25 \mathrm{~V}$ | C 6 | $100 \mu \mathrm{~F} 50 \mathrm{~V}$ |

Semi-conductors:

| Tr1 OC45 | D1 OA202 | D4 GJ7-M (AEI) |
| :--- | :--- | :--- |
| Tr2 OC81 | D2 | OA202 |
| Tr3 OC81 | D3 | GJ7-M (AEI) | D5 GJ7-M (AEI) $^{\text {D6 }}$ GJ7-M (AEI)

Miscellaneous:
S1 2-pole 3-way rotary. S2 on-off toggle. T1 Transformer Type T1079 (Ardente-Henrys Radio) T2 Transformer Type MT-9 (Osmor-Henrys Radio) Speaker (see text) Material for paddle. Relay (see text).
centrally lengthwise provide two blades; hacksaw blade may also be used. Finger-plates of thin perspex are easily fixed at the operating end of the paddle.

The relay used is obviously of considerable importance, but STC sealed types with 520 ohm coils work well and coil resistances up to 820 ohms should also be satisfactory provided the relay is light and reasonably fast in operation. Types STC4190GD and 4189GD available from Henry's Radio should be suitable, but if any doubt exists or where the constructor wishes to try using a specimen already to hand a small pre-construction "mock-up"' may prove enlightening. Capacitor values may have to be modified slightly


- Fig. 2. Details of the construction of the paddle bracket.

FFig. 3. Circuit of the suggested power supply unit.


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depending upon other components used, but wide variations from values specified are considered unlikely.

## Power Supply

The supply potential should not be less than 9 volts d.c. and the current drain demanded will approximate to 20 mA during operation. A battery is the obvious source of power but use of an adjustableoutput mains driven unit may be considered; a suitable circuit is given in Fig. 3. Here the mainsisolating transformer selected measures but $1 \frac{1}{8} \times$ $1 \frac{1}{2} \times 1 \frac{1}{4}$ in which means that the entire p.s.u. can be built into a space no greater than that needed for a type PP9 battery; this unit will also be found useful in other workshop connections.

By using the whole of the transformer secondary winding in conjunction with four rectifiers bridgeconnected as shown up to 12 volts d.c. at 35 mA can be taken or 15 volts at 10 mA ; the transformer is rated to 80 mA .

## Conclusion

Housing the completed keyer in a box made from aluminium sheet is desirable and if the battery is to be used for powering this can be accommodated within the box. If the mains-derived power supply is adopted it may be found more convenient to keep the p.s.u. as an item separate to the keyer.

Due to the integral audio monitor there is no need initially to switch on the transmitter to get the feel of the keyer-in fact, your friends will thank you for not doing so! The keyer does take a little while to get used to after operating with a conventional key and it therefore is wise to spend some time practising. When the various controls and the paddle have been adjusted to suit individual fingering subsequent "on the air" operation should prove both pleasurable and relaxing, but if you do feel the urge to speed your sending do not forget the chap at the other end!

## CORRIGENDA

## October 1970. Mini-Organ.

Fig. 2. Connection between VR12 and VR11 should be extended to dotted arrow 11. Similarly the connection between V13 and V12 should be extended to dotted arrow 12.

Fig. 3. The capacitor between rails $M$ and $S$ should be marked C6 and not C4. Transistor $\operatorname{Tr} 4$ connection to rail M is the collector, to rail O the emitter and to rail T the base.

## October 1970. Transistor and Diode Tester.

In the components list on page 431 the specifications for switches S1 and S2 should be transposed.

September 19\%0. Guitar Amplifier, Part 1. To the components list add R47, $150 \mathrm{k} \Omega$ and amend Henelec mains power unit reference number to Type MU442, (not 422).

Finish off that Equipment-continued from page p. 848
Any screwheads that must appear on the panel should be plated ones. In general tighten up nuts from the back of the panel rather than using a screwdriver on the screw. The slot may be damaged or, far worse, the screwdriver may slip and ruin the panel.

These precautions are doubly necessary when using Letraset as a panel finish as it is easily pulled up when tightening screws or nuts from the front. Always use a flat washer under a lock nut.

As far as possible knobs and dials should match in size and colour and again a little extra expense in this direction is well worth while. If the equipment is mains operated don't forget to fit an indicator lamp to the front panel because apart from the pleasant effect it prevents the equipment being left switched on inadvertently.

It is also a good idea to fit an on-off switch to each separate piece of equipment rather than to have a loose plug that has to be picked up and plugged into the mains every time the equipment is wanted.

If one is likely to be a prolific producer of gear it is advisable to settle on a particular make or design of cabinet and colour scheme so that eventually the units will blend together instead of finishing up with a motley collection of cabinets of varying sizes and colours.

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| OB2 | . 30 | 6 F 18 | . 45 | 12A6 | . 63 | 30 FL 12 | . 80 | DH77 | . 80 | ECF80 | 33 | EM34 | 90 |
| 1.A7GT | . 37 | $6 \mathrm{~F}^{2} 23$ | . 72 | $12 \mathrm{AC6}$ | . 40 | 30FL14 | . 73 | DK40 | . 55 | ECF82 | 33 | EM880 | 8 |
| 1H5GT | . 35 | 6F24 | . 68 | 12 AD 6 | . 40 | 30 L 1 | . 32 | DK92 | . 43 | ECF86 | 65 | EM81 | 42 |
| 1 L 4 | . 18 | 6F25 | . 65 | 12AE6 | . 48 | 30 L 15 | . 64 | DK96 | . 37 | ECF8042 | 10 | EM84 | 34 |
| 1N5GT | . 39 | 6 F 28 | . 70 | 12AT6 | . 23 | 30 L 17 | . 78 | DL96 | . 37 | ECH21 | 68 | EM87 | 38 |
| 1 R 5 | . 28 | 6H6GI | . 15 | 12AT7 | . 19 | 30 P 4 MR | . 88 | DM70 | . 30 | ECH35 | 29 | EY51 | 37 |
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| 185 | . 22 | 6 J 6 | . 18 | 12AU7 | . 23 | 30P19/ |  | DW4/ |  | ECH81 | 29 | EY83 | . 55 |
| 1 U 4 | . 29 | 6J7c | 24 | 12AV6 | . 28 | $30 \mathrm{P4}$ | . 60 | 350 | . 38 | ECH83 | . 40 | EY84 | 50 |
| 105 | . 48 | 6J7GT | . 38 | $12 \mathrm{AX7}$ | . 23 | 30PL1 | . 69 | DW41 |  | ECH84 | . 38 | EY86/7 | . 33 |
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| $3 \mathrm{Q5G}$ ' | . 35 | 6L6GT | . 39 | 12E1 | . 85 | 30PL15 | . 98 | E80F | 1.20 | ECL84 | 60 | EZ41 | 43 |
| 354 | . 29 | 6LTGT | . 63 | 12J7GT | . 33 | 35L6GT | . 44 | E83F | 1.20 | ECL85 | . 55 | EZ80 | 28 |
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| $5 \mathrm{Z4G}$ | .35 | 6Q7 | . 43 | 12SA7GI |  | 50 L 6 GT | 45 | E1148 | . 58 | EF37A |  | FW4/ |  |
| $6 / 30 \mathrm{LZ}$ | . 58 | ${ }^{607}{ }^{\text {a }}$ | . 30 |  | 40 | 72 | .33 | EAAOO | . 88 | EF39 | 40 | GZ30 |  |
| 6A8G | . 38 | 6R7 | . 55 | 12SC7 | .35 | 85 A 2 | 43 | EA76 | . 83 | EF41 | .50 | GZ33 |  |
| 6AC7 | . 15 | 6R7G | .35 | 12SG7 | .23 | 887 | . 50 | EABC80 | . 38 | EF42 | .38 | GZ34 |  |
| 6AG5 | . 2 | ${ }^{685 A 7 G T}$ | . 35 | 128J7 | . 15 | AC | . 50 | EAF42 | .50 | EF54 | .88 | GZ37 | 75 |
| 6 6AL5 | . 12 | 6SC7GT | 38 | 12SK7 | . 24 |  | 98 | EB34 | .20 | EE73 | . 38 | HABC | 45 |
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| 6AT6 | . 20 | 6SH7 | . 53 |  | 50 | DD | 98 | EB91 | . 12 | EF83 | . 48 | HL41 | . 98 |
| 6AU6 | . 25 | 6857 | . 35 | 1487 | 1.15 | AC6PPE |  | EBC41 | 48 | EF85 | 29 | HL42D | . 50 |
| GAV6 | . 30 | 6SK7GT | 23 | 19AQ5 | 24 | AC/PEN( |  | EBC81 | . 33 | EF86 | 32 | HN309 | . 38 |
| 6B8C | . 13 | 6SQ7GT | . 18 | 20 F 2 | . 70 |  | 98 | EBC91 | . 30 | EF89 | . 25 | HVR2 |  |
| 6BA6 | . 23 | 6V6G | 18 | 20 L 1 | . 98 | AC/TH1 | 50 | EBF80 | . 34 | EF91 | 17 | HVR2A |  |
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| ${ }_{68 \mathrm{BR} 7}^{68}$ | . 79 | $7 \mathrm{B7}$ | . 30 |  | . 29 | AZ31 AZ41 | . 48 | EC88 | . 35 | EH90 | . 38 | KTW61 | . 63 |
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| 1509 | . 73 | 10 Cl | 1.25 | 25Z5 | . 40 | DAC32 | . 35 | ECC82 | . 23 | EL41 | 55 |  |  |
| 6CD6G | 1.15 | 10 C 2 | . 50 | 2076G | . 43 | DAF91 | . 22 | ECC83 | . 23 | EL42 | 53 | All va | s |
| 6 CH 6 | . 38 | 10 F 1 | .75 | 30 Cl | . 30 | DAF9 | . 35 | ECC84 | . 30 | EL81 | . 50 | Cash | her |
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| 6F6G | . 25 | 10P14 | 1.10 | 30FL4 | . 64 | DH63 | . 30 | ECC80 | 8 | EL9] | 28 | imite | nd |

|


Number 16

## Operational Amplifiers in Audio Circuits

OPERATIONAL amplifiers have not figured significantly in these notes to date, despite the fact that, in terms of numbers produced, these are by far the most popular linear i.c.'s. They have not figured, though, for the very good reason that their applications lie mainly in the field of industry and instrumentation; indeed the phrase comes from the origin of this type of device, in the simulation of mathematical operations in analogue computing.

The primary feature required in an operational amplifier is the ability to cope with input signals varying in voltage only slowly (i.e. pseudo direct currents) representing for example the temperature or pressure at which a chemical reaction takes place. Interstage coupling with capacitors is therefore impossible since these have a frequency-dependent impedance, so the designer must resort to direct coupling throughout to achieve a fixed gain with signals ranging from virtually d.c. to (typically) 1 MHz .

Further requirements are a low drift (good longterm stability in gain and voltage levels) with a high input impedance and a low output impedance, and high voltage gain. These features ensure that the circuit imposes only a small load on the instrument or transducer driving it, while output is only slightly affected by the drain of the following stage.

The mathematical operations of addition and subtraction, integration or differentiation are achieved by resistive or capacitative feedback, which also limits the gain of the amplifier from the very high "open loop" figure to that required for the specific application.

Two input terminals are available, the negative (to which the feedback connection is made) which produces an output signal of opposite polarity to the input (corresponding to multiplication by -1 or phase shift of $180^{\circ}$ depending on the particular simulation) and the positive which secures an output similar to the input in sign and phase.
Fig. 1 shows an operational amplifier in a simple simulation. An extended discussion of the theory of the operational amplifier, centred on the type 709 which is the standard model monolithic circuit, was given in the August 1970 edition of our companion journal, "Practical Television".

Several of the units previously treated in these pages were adapted from industrial applications, and the same potential exists for the operational amplifier. The characteristics of the circuit make


Fig. 1: Operational amplifier as used in mathematical simulation of expression shown. Input impedance $=R_{\text {IN }}$ since feedback produces a "virtual earth" at the input terminal.
it ideal for audio and video amplification, with its wide bandwidth and stability. With additional power transistors, a first rate hi-fi unit can be assembled at low cost and with minimal possibilities for error. Indeed the power transistors can be brought within the feedback loop of the operational amplifier, so that crossover problems, and other sources of distortion and noise are automatically compensated.

Complementary transistors allow the principle of


Fig. $2: 4$ watt mono amplifier, using the MC1533 (TO-5). Numbers in brackets refer to economy style MC1433P, electrically identical, in 14 pin package.

direct coupling to extend to the output stage, with the result that the amplifier has a fixed gain from d.c. to above 20 kHz (there is no low frequency cut off). At the same time the theory of the operational amplifier can be applied to the complete circuit to determine its gain, while mixing facilities are available at the input as in addition simulation.

Two applications of these ideas are illustratedone a domestic sound system, Fig. 2, the other for p.a. work, Fig. 3. It will be noted that a "split" power supply is used, corresponding to industrial practice. This has the advantage that all signals are referred to earth, and the load is directly connected from the amplifier output to earth without the necessity
of a large blocking capacitor, a not unusual source of trouble in more conventional schemes.
For stereo operation, there is an alternative to straightforward duplication of one of the amplifiers illustrated. Readers may recall that in the August 1970 issue a Motorola unit MC1303 was introduced as a stereo preamplifier, and it was mentioned that it was a development of the industrial unit MC1535. In fact this is a dual operational amplifier, and the remarks given above apply to each of the sections independently. Given the pin connections (Fig. 4) it should be a challenge to the reader to design a stereo system along the lines indicated for the mono amplifiers of Fig. 2 or Fig. 3.


Fig. 3 : (left) A 20 W mono hi-ff amplifier suitable for p.a. work. Fig. 4 : (above) The connections for the MC1303 a dual operational amplifier suitable for stereo work.

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## 



## MONTHLY

 NEWS FOR OX LISTENERSHE onset of winter has seen the usual increase in activity and a consequent increase in the number of reports to this column. The first comes from:
C. R. S. Stacey of Tunbridge Wells who tuned his Lafayette HA-700 receiver and Joystick antenna to hear:
4875 S. Africa, Domestic Sce., Afrikaans at 1924
4890 Radiodiff. du Senegal in French at 1900
4915 Voice of Kenya, Home Sce., Swahili at 1820
6040 Voice of the Coast (Sharjah), Arabic at 1400
6045 R.R. Indonesia in Indonesian at 1521-1537
6045 Greek Armed Forces Information Service at 1530
6065 Radio Voice of the Gospel, Amharic at 1630
9770 Vladivostock, Ext. Sce. of R. Moscow at 1500
15280 Voice of Malaysia, Indonesian at 1543-1553
Jeffrey Malina of London, N.4, is a new reporter who used his Skyrover Mark II receiver and 104-foot long-wire antenna to hear:
5920 Kiev, Ukrainian S.S.R., noted at 1930
7065 Tirana, Albania, from 2130 to 2145
9009 Israel B.S. from 2045 to 2100 and 2115 to 2130
9625 Israel B.S. from 2045 to 2130
9520 Trans World Radio noted at 1700
9912 All India Radio from 2045 to 2115
17715 Radio Havana, Cuba, at 2105 to 2140
17725 Radio Cairo, U.A.R., from 1415 to 1430
17870 Radio Ghana noted from 1400 to 1410
Regular reporter John Trewick of London Colney has heard the following stations:
9635 Baghdad, Iraq, news in English at 1935
11790 ABC, Australia, in Vietnamese at 2130
11845 Trans World Radio, Bonaire, testing at 2140
11855 Saudi Arabia with news in English at 1950
15080 All India Radio with English to Africa at 1900 15320 ABC, Australia, with "Mailbag" at 2115.

Robert Cork of Basingstoke has sent in another report which included:
7065 Radio Tirana, Albania, at 1845
9625 Israel B.S. at 1130 with religious programme
9665 Radio Malaysia from 1100 to 1200
11805 Radio Globo from 2215 to 2225
Nigel Milner of Sutton Coldfield used his Invicta 5 -valve domestic equipment to hear some interesting stations including:
6095 Radio Baghdad, Iraq, in English at 2020
9009 Radio Israel in English at 2015
11915 HCJB, Quito, Ecuador, in English at 0430
15250 RSA, S. Africa, with news in English at 1900
15405 Radio Kuwait, Top Twenty Show at 1700
Patrick Hendrick of Wellington in South Africa has built various pieces of equipment as described in P.W., including the Modular 3-band s.w. receiver. This equipment and three 50 -foot long-wires enabled Patrick to hear the following:
3925 Australian B.C., VLK3 at 0740

# THE BROADCAST BANDS Malcolm Connah 

## Frequencies in kHz - Times in GMT

## 4896 Australian B.C., VLT4 at 0740

6250 Lourenço Marques, Mozambique, at 1830
7205 Voice of America relay at 1745
11760 B.B.C. noted in English at 2145
15125 Voice of Free China, BED60, noted at 1705
15140 B.B.C. noted at 2145
17845 Radio New York World-Wide at 1710
Nicholas Hall of Cheadle has an AR88D receiver, a BC221 frequency meter and a 150 -foot long-wire antenna, which enable him to hear:
4825 Radio Ashkabad, USSR, at 1710
4865 Ponta Delgada, Azores, at 1845 in Portuguese
4940 Radio Kiev, Ukrainian S.S.R., at 1540
5040 Radio Tbilisi, USSR, at 1615
15020 Radio Hanoi, Vietnam, at 1300
17705 Radio Havana, Cuba, at 2015
21545 RSA, S. Africa, to N. Zealand at 0800 to 0850
Derek Hart of Lancaster has an Astrad Auriga receiver and an excerpt from his $\log$ reads as follows:
6065 Seoul, Korea, with news in English at 2115
9390 Radio Tirana, Albania, in English at 1835
9550 Radio Finland, in English at 1000
15365 RSA, S. Africa, in English at 1000
15405 Radio Kuwait, documentary at 1730
17945 Karachi, Pakistan, news in English at 1515
21460 HCJB, Quito, Ecuador, in English at 1945

## News

Regular reporter Roy Patrick of Derby has sent in some more items of news. Many thanks, Roy, for another excellent contribution.

ECUADOR, HCJB, Quito is now using the frequency of 15300 instead of 15425 for the European Service. The best frequency for reception in this country is 21460 .

LIBERIA, ELWA, Liberia has been logged on the new frequency of 15170 at 2000 . The programme heard was in French.

GHANA, Radio Accra, Ghana has been heard at 1445 on 21545 with a programme in English.

UNITED ARAB REPUBLIC, The Voice of Africa from Cairo has been heard on 17725 in English from 2030 to 2155.

UNITED STATES OF AMERICA, WNYW, Radio New York World-Wide has been reported to be up for sale by several DX publications. According to a station announcement the station is very definitely not for sale.

All reports, which should be in frequency order, must arrive by the 17 th of the month. They should be addressed to me at 5 Ranelagh Gardens, Cranbrook, Ilford, Essex.


ACH year begins with a new crop of resolutions all made, at the time, in good faith. How about making some of them a reality this year? That crystal frequency marker which would enable the receiver dial to be made far more meaningful is an example of planning something concrete and sticking to it.

Others might like to make their resolutions, or plans, for later on in the year. One can usefully employ the cold winter months planning some antennas which might be tried. How about a ground plane, or even a vertical dipole. Neither antenna is difficult to make and plans as to construction and location could be made so that the antenna can materialise with the minimum of effort once the warmer weather arrives.

For those who want to get a bit more out of their hobby, a world map plus a good atlas can work wonders. The world map goes on the shack wall and small coloured pins may be pushed in when stations are either worked or logged. When a station gives his precise QTH, it is interesting to check in the atlas to see just where some of these funny-sounding places really are. This scheme does a number of things besides giving added interest. It will build up quite a good knowledge of geography without really trying and will assist in associating the callsigns of countries with their exact location.

Don't forget that receivers, like motor cars, need servicing. A check up and re-alignment can work wonders and simple accessories can be a great help. A simple oscilloscope hooked on to the first i.f. transformer can give a very good indication of what the modulation of the incoming signal is like and will show any whiskers on a c.w. signal.

None of the above suggestions are obscure or ingenious. However, readers might care to sit down and think how they will be getting the most out of their hobby this year. Perhaps you can think of an idea better suited to your own particular needs.

The l.f. bands have given a clear indication this past month that things are happening for those who can be bothered to listen. Anyone can $\log$ DX on 14 and 21 MHz but it takes a definite skill on the lower frequencies and not just a good receiver.
Steve Champion (Herts.), reports the 7 MHz at 0730 is a good time to harken to VK types and the like with little QRM to make things difficult. Both VK3OZ and VK3ZL were logged at 5 and 7 on s.s.b
Down at Weymouth, Les Barnes has been stoking up his CR70A on 3.5 MHz . Rewards from-CR4BC, K5PFL, LXIDAC, W2TF, ZL4IE, 9H1BW, 9V1PP.
Tim Thornton (Berks.), CR70A, 132ft. end fed also reckons that conditions are excellent on eighty metres. Extract from his s.s.b. log reads-EA4JL, FB8XX, HB0XFB, VE1IE, VE1AAW ( 5 and 9 plus at 0640), VO1FG, WA8ROJ (c.w.), YT2NFJ, ZB2A, ZM4PG, OH1PS/4X, 5Z4KL, 6W8DY, 9X5PB.
Still on eighty, Charles Lewis (Flint.), is also an 1.f. addict. Armed with W.Ş. 19 set Mk.III he hooked KCVR and VS6DO on a.m. and VK6DF on s.s.b.

# THE AMATEUR BANDS David Gibson, G3JDG 

## Frequencies in kHz - Times in GMT

P. Baldwin (Northants), homebrew receiver, 67 ft . end fed, 80 s.s.b.-CT1NK, HC4BS, K3UZE, TF5TP, VE1AA, VE1IE, VE8RX, VO1BV, W2HCW, XE1KB, ZL4IE, ZL4N̦J, ZS5XA, 9F3USA.
At Maldon in Essex, Paul Fereyer has a Marconi No. 9 set and a 100 ft . wire. Goodies on eighty include - CN8HD, EP2DX, JA5CQ, JX8XF, K8HCL, PJ0DX, VE3TX, VOIFX, VP1WMU, VP2VP, VP7BW, VP9DX, VS6DO, VS9MB, XE1KB, ZL4IE, ZL4JW, ZL4KE, ZM1AIX, 4X4UF, 6W8DY, 9J2BG.

Not up but down. Stephen Davies (S. Wales), B40, 250 ft . long, long wire, reports signals on Topband from-EI9ONE, OK1AA, OK1AKN, OK1ATP, PO0SE, ZC4RB.
If you thought we'd finished with eighty, you're wrong. John Moxham (Somerset), SR200, 67ft. end fed plus a.t.u. says it's been a great month for 3.5MHz. His log reads-FB8XX, JX8YN, K4MMY, OY7JD, OY9LV, PJ7JC, TJ1AW, TZ2AB, VE1AA, VE8YL, (Resolution Is.), VS6DO, VS9MB, VU2BEO, WA5AHB, XE1KB, ZB2A, 4S7AB, 4U1ITU, 4X4KT, 4Z4HF, 6W8DY, 9F3USA.
On the h.f. bands things have been pretty lively adding up to a very nice six-band month as regards activity. Twenty metres gave John Morrie (Lancs.) AX2ZQ, OA7KP, VP80J and VR6GA. Up on 21 MHz , Mark Wallace (Lincs.) grabbed himself his first Australian call from VK2FU.
Philip Beeson (Staffs.) HA500 and 67ft. end fed pricked up his ears to signals from JY1 on 14,300 s.s.b., while a listen on 28 MHz raised calls like CR7FO, W5KDI and YS2ACB. S. Wainwright (Lancs.) heard CR7IZ, EP2FB, FR7ZN, JY1, KP4ADY, LU5DL, MP4BHL, TJ1AZ and 8P6TCA on ten metres but makes no mention of what receiver or aerial. (Perhaps he's psychic?).

An ancient Ferguson domestic receiver is ye sad lot of P. Harris bought second hand in 1946 (the set -not P. Harris). Trouble is, when he connects his homebrew preselector and an antenna he hears things on twenty metres, like-CE3TP, CR6FR, FB8XX, FR7ZN, JAINAW, KR6TQ, ZM4BO, 8P6AZ and 9K2AM.
Listening in the the school holidays is the $\sin$ of Paul Austin (unpronounceable place in Monmouthshire). The CR100/2 plus dipole on twenty aided and abetted with-AP5HQ, AX7GC, CR6IK, FO8BY, JA6JBR, KH6FF, LU1ZE, PZ1BK, SU1MA, ZL3VV, ZM3US, 9Q5AS.

An 8 -over-8 at 30 ft., m.o.s.f.e.t. converter and Trio 9R59DE together with John Haig all live in Hitchin, Herts. Combined effort for two metres includesG8EGS, G8BZR/P (Rutland), G3TUW, G8ATK, G8BXX, G3DAH, G8BCL, G3BGH, G8BTX, G8CEX,

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[^2]

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We knew about RMS--but, of course-of IHF, Peak IHF and all the twists of Music Power, but WFT was a new one. Why should Messrs Blank use that particular rating, we enquired. He snorted: "Don't they always? I've never known a bigger bunch of Wishful Thinkers."

Abbreviations are O.K., just so long as we all talk the same language. According to the British Association, there are 88 ways in our dialect of saying "lefthanded." In the radio trade, we get used to LH. We use it as we use the red plug for the right channel, and the copper wire for earth.

Except that the plugs supplied will be blue and yellow, or both grey, and the twin cable will be twin copper, good for guessers. There was some hope that decimalisation would improve all this, but by the time we have sorted out our dodecacahedral from our equilateral curved heptagon coins, we haven't the energy to spare.

As for the chap who, with no foreknowledge, has to work out those audio symbols-Heaven help him. A gramophone pickup sign is simple: it even begins to look like an old-fashioned playing head (which is more than some modern ones do). That dinky aerial symbol is pretty meaningful. But when we get inverted top hats, buckets on their sides and pairs of wavy lines, even, bless us, a lightning flash, we hardly know which button to press.

If in difficulty, call HELP. That is, Henry's Enquiry Liaison Panel.

Books reviewed on this page are normally obtainable through any retail bookshop. In this instance, the information printed in heavy type should be quoted.

## TUNERS AND AMPLIFIERS By John Earl <br> Published by Fountain Press Ltd., 46-47 Chancery Lane, London W.C.2. 187 pages, $8 \frac{1}{2} \times 5 \frac{1}{2}$ inches. Hard covers Price 42s.

HAIL and welcome to the first of a new series of books on radio, television, audio and allied subjects. This venture by Fountain Press deserves success, if only for the meticulous way the series has been planned and the care with which preparation and layout have been made.

The project sounds simple enough-commission a top author in each branch of the subject and let him loose within a set format; edit his work and embellish it with tasteful illustration, then sit back and wait for the sales. We all know that it is never so simple, and if Mr. Earl's book reads effortlessly we may be sure it is the result of great diligence and not a little expertise.

It is certainly comprehensive. The field of amplifiers and tuners has grown apace in recent times; new devices have come along; new techniques flourish; we are even blessed with new basic materials. Mr. Earl has performed a much-needed service for the technician and the enthusiast in collating, summarising, discussing and in many cases explaining in admirable detail the modern tuner and amplifier.

He states his aims at the outset, to have 'maximum appeal not only to the user of high quality audio equipment but also to the person who is about to start out on the hi-fi adventure.' The subtle technicalities do not bog us down, there are, as Mr. Earl says, a number of books that fill this role. In eight chapters, he roves the difficult country of semiconductors in modern equipment, the ups and downs of tuner circuitry, the controversial ground of amplifiers, with many practical examples, touches upon aerials, systems, the addition of a cartridge, power supplies, headphone and speaker matching and has even found space to add an excellent chapter on specifications.

The writing is concise, the illustrations well chosen and clear and the information throughout as up-todate as a new-minted coin. Reading through this book, we come up against the very personal approach quite unexpectedly and if it teaches us to rethink, so much the better. Take for example Mr. Earl's remarks on 'Dynamic Range' in Chapter Four: exploding the wattage myth he comments, 'To reproduce 70 acoustic watts in full-force from a pair of speakers of $5 \%$ efficiency would require 350 electric watts in each channel . . . without distortion.' There are watts and watts, and it is not said often enough!

This is not specifically a 'hi-fi' book. It caters for the chap who can only afford mid-fi equipment as well as the true audiophile. It is, in fact, one of the few books I have read which could help to turn the one into the other. It earns a welcome
place on my bookshelf, from where I know it is going to be withdrawn many times for some reference or other to be checked. I shall probably be thankful for the index, which covers well over four pages. I shall certanly be grateful for Mr. Earl's easy style and look forward to seeing later contributions by him in what promises to be a useful and entertaining series of technical books.-AWH.

## TAPE RECORDER SERVICING GUIDE By Robert G. Middleton Published by Foulsham-Sams 96 pages, $11 \times 8 \frac{5}{8}$ in. Price 35s.

IHAVE only one serious quibble about this book: its size. Large of page and thin of stature, it is more suitable for the coffee table than the bookshelf, and books laid on tables in my workshop tend to be used as wedges and props.

Mr. Middleton's latest work-the latest of many, for RGM is the American equivalent of our own Gordon J. King-is worthy of more than a casual dismissal. He has managed to compress a wide field of specialised service work into seven chapters, of which all but the first are strictly practical and to the point.

This first chapter lays the ground quite efficiently, describing the general principles of tape recording, with much emphasis on basic magnetism, the theory of bias and mechanical drive systems. For the reader unacquainted with tape recording, this chapter is necessary, for the more experienced audiophile it nevertheless offers an interesting and quite personal review of the subject.

After this we have the tape recorder described in detail by first dealing with maintenance and testing, then adjustments and minor repairs, and finally, in chapters 4 and 5, the troubles that can beset the machine's transport and recording circuitry. After this, chapters on relay troubles and those that can present problems with associated equipment underline the points and fill in a few of the gaps. But it is the heart of this book that is the vital part for the student of recording.

I would have liked more definite examples of recording equipment: nowhere is apparatus named; and yet valuable space is wasted on parts lists, complete with computer programme parts numbers. Some of the circuitry could have been brought nearer the present day, also, for this is a copyright 1970 work, and not a rewrite polished up for the benefit of we poor Limeys.

For those who have trouble wrestling with decibels allow me to commend the section on microphones, where one of the neatest explanations of relative levels, standard values and the importance of similar load values that I have ever seen is most admirably expressed. As, for that matter, is the whole book, making interesting and informative reading that is thoroughly recommended.-HWH.

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[^3]

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clear dial, size $4 \frac{3}{7} \times 3$, which can be set in minutes up clear dial, size $4 \times 3$, which can be set in minutes up
to 1 hour. After preset period the bell rings. Ideal for processing, a memory jogger or, by adding simple lever, would operate micro-switch 22/6.


## THE FULL-FI STEREO SIX



The amplifier
You will sensation of the year You will be amazed at the fullness of reproduction and at the added qualities your records or tuner will reproduce. Built styled in simulated teak finished
to blend with modern furnishings, this amplifier nae an integrated solid state circuit with an output power of 6 watts R.M.S. split over the two channels. The amplifier is ideal for use with normal pick-ups and tuners, it has a double wound mains transformer and ganged volume and tone controls-also switching for Mono to Stereo, tuner or pick-up. Other controls include 'treble lift and cut", "balance", and separate mains on/ofit switch. UNREPEATABIE PRTCR is $\$ 8.19 .6$ plus $7 / 6$ post and insurance.

| No of Poles |  |  | standa contact washer 4 way |  |  | wafer pindle <br> 8 way | silver-p <br> ${ }^{\prime \prime}$. long <br> 9 way | lated -with <br> 10 way |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 pole | 6/6 | 6/6 | 6/6 | 6/6 | 6/6 | 6/6 | 6/6 | 6/6 | 6/6 |
| 2 poles | 6/6 | 6/6 | 6/6 | 6/6 | 6/6 | 6/6 | 8/6 | 10/8 | 10/6 |
| 3 poles | 6/6 | 6/6 | 6/6 | 6/6 | 10/6 | 10/6 | 10/6 | 14/6 | 14/6 |
| 4 poles | 6/6 | 6/6 | 6/6 | 10/6 | 10/6 | 10/6 | $10 / 6$ | 18/6 | 18/6 |
| 5 poles | 6/6 | 6/6 | 10/6 | $10 / 6$ | 14/6 | 14/6 | 14/8 | 28/6 | 22/6 |
| 6 poles | 6/6 | 10/6 | 10/6 | 10/6 | 14/6 | 14/6 | 14/6 | 28/6 | 28/6 |
| 7 poles | 10/6 | 10/6 | 10/6 | 14/6 | 18/6 | 18/6 | $18 / 6$ | 80/6 | 30/8 |
| 8 poles | 10/6 | 10/6 | 10/6 | $14 / 6$ | 18/6 | 18/6 | 18/6 | 34/8 | 34/6 |
| 9 poles | 10/6 | 10/6 | 14/6 | 14/6 | 22/6 | 22/6 | 22/6 | 38/6 | 38/6 |
| 10 poles | 10/6 | 10/6 | 14/6 | 18/6 | 22/6 | 22/6 | 22/6 | 42/6 | 42/6 |
| 11 poles | 10/6 | 14/6 | 14/6 | 18/6 | 26/6 | 28/6 | 26/6 | 46/6 | 46/6 |
| 12 poles | 10/6 | 14/6 | 14/8 | 18/8 | 28/6 | 26/6 | 28/8 | 50/6 | 50/6 |

## SPARTAN Portable

## RADIO

Long and medium wave, 7 transistor, size 6 in. $x$ than nsual speaker riving very good tone. Built-in ferrite aerial and telescopic aerial for distant
 stations. A real bargain complete with leather case, carry sing, earplug and case. 83.15.0. plus $5 / *$ post and ins.

## MULTI-SPEED MOTOR <br> Replacement in many well known food mixers. Six speeds are available. 500,850 and 1100 r.p.m. from either or both of the nylon sockets (Where the beaters of the food mixers normally gos and $8,000,12,000$ and 15,000 <br>  

r.p.m. (ideal polishing speeds) from the main drive shaft. Very powerfuland useful motor size appox. ghaft. Very powerful and useful motor size appox.
2 in. diameter, 5 in. long. Price $17 / 6$ plus $4 / 6 \mathrm{p}$. ins.

## MAINS OPERATED CONTACTOR

 $220 / 240 \mathrm{v}$. 50 cycle solenoid with laminated core so very silent in operation. Closes 4 circuits each rated at 10 amps. Extremely well made by a German Electrical Company 19/6 each.

## DOUBLE ENDED MAINS MOTOR

 On feet with holes for screwdown flxing. 'To drive models, oven, blower heater, etc. 10/each, plus $3 / 6$ post and insurance, 6 or more post free.0.005 mFd TUNING CO NDENSER
Proved design, ideal for straight or - reflex circuits $2 / 6$ each, $24 /-$ doz.


Where postage is not stated then orders over $£ 3$ are post free. Below $£ 3$ add $2 / 9$. Semi-conductors add $1 /$-post. Over $£ 1$ post free. S.A.E. with enquiries please.

## ELECTRONICS (CROYDON) LTD

Dept PW 266 London Road, Croydon CRO-2TH Also 102/3 Tamworth Road, Croydon

MW/LW TRANSISTOR RADIO
Complete set of parta for this MW/LW Transistor Radio. Originally sold by Forldfamous company as a set for more than $£ 17$.
 + 2 OC81. Also works off car-werial. Full instruction book $2 / 6$, free with parts. State preference red, blue or tan and give alternative. Requires PP7 battery, price $4 / 6$ extra. fully built $+4 / 6$ post. Kix (see previous A few left of Super Six Kit (see previous igsues) at e4.5.0 ( $+4 / 6$ post).
9 volt $\overline{6}$ transistor superhet using 3HF AA17 in RF and IF stages powered by a two stage AF amplifier using OC81D. $2 \times 1$ © $\mathrm{CB1}$.
Output approx. 500 MW .

## STEREO AMPLIFIER type HV- $2 \times 3$ Watts

Fully built. On off sep. vol. and tone each ohannel 12 speakers, double wound main trans: fixing flanges speakers, double wound main trans: ixing tanges ${ }^{£ 6} 6.10 .0$ ( $8 /-\mathrm{p} . \& \mathrm{p} ., 200-250 \mathrm{~V}$. A.C.
Superior version - separate base. treble $\&$ vol each channel with mono/stereo switch 15/. extra.

## MULLARD STEREO "DO IT YOURSELF"

Pre-amplifier EP9001 88.2 .0 . 2 amps. EP9000 $£ 2.18 .0$ each. Power Supply EP9002 £4.12.0. Control panel $\mathbf{2 2 . 1 0 . 0}$ or the lot for $£ 15.10 .0$ post paid. Booklet $5 / . \cdot$ ( 6 d post)


MONO GRAM CHASSIS 3 WATT
3 Wave band long-med-short. Gram., $200-250 \mathrm{~V}$ A.C Ferrite aerial. Chassis $13 \times 7$ x Ein. Dial $13 \times 4$ in.
Double wound mains transformer 5 valves ECH81, Double wound mains transformer 5 valves ECH81,
EF89, EBC81, EL84, EZ80. Price $£ 10.18 .8$.
$(7 / 6$ p, \& p.) Output trans. for 3 -ohm speaker (7/6 p, \& p.) Output trans. for 3 -ohm speaker

NEW TAPE AMPLIFIER for 4 track B.s.R. Deck TD2. Mains and output Trans. $10^{\prime \prime} \times 6^{\prime \prime} \times 4^{\prime \prime}$ overall. Rect; ECC88, EL84. Mike and Gram Inputs. © 7.5 .0 . (10/-p. \& p.) For 3 -ohm speaker. (For $200-250 \mathrm{~V}$. A.C.)

$1 \frac{1}{2}$ W MAINS GRAMOPHONE AMPLIFIER
EZ80, ECL82, O.P. Transformer (3 ohm). Vol./On-ofi and Tone Control. Double wound mains transformer. $2 \frac{5}{6} \times 2 \frac{1}{8} \times 2 \frac{1}{4}$ in. separate but wired to chassis, $4 \times 2 \frac{1}{2} \times 4 \frac{8}{4} \mathrm{in}$. over valves $67 / 6$ including 万̂in. Speaker ( $55 /$ - less speaker). (Post paid.) (Two less $10 \%$ ).

SINCLAIR AMPS \& PRE-AMPS IN STOCK. RETURN OF POST SERVICE.

MAINS TRANSFORMERS (200-250V AC INPUT)
Prstage. Postage in braokets.
Out 250 V at 50 mA and $6-3 \mathrm{~V}$ at $11 \mathrm{~A} .10 /-(2 / 6)$.
$6.3 V$ at 21 A $7 / 6$ ( $8 /-$ )
22 V at $1 \mathrm{~A}+6.3 \mathrm{~V}$ at 2 A and 250 V at $50 \mathrm{~mA} 15 /-(3 / 6)$. 250 V at $40 \mathrm{~mA}, 24 \mathrm{~V}$ at 1 A and 6.3 V at $1 \frac{1}{\mathrm{~A}} 12 / 6$ (2/6).
250 V at $50 \mathrm{~mA}, 22 \mathrm{~V}$ at 1 A and 6.3 V at $1 \frac{1}{1} \mathrm{~A} 17 / 6(3 / 6)$. 90 V at 20 mA and 1.4 V at $250 \mathrm{~mA} 10 /-(2 / 6)$.
 dozen post free. Metal with grey-green finish.

## NEW F.M. TUNER

Range 87 to 107 MHz Attractively finished metal container with cast front escutcheon. Case size Rectifier and Yalves ECC85, EF89, EF80, ECC82 as cathode follower. EM84 tuning indicator, $2 x$
AF117 and 2 diodes. A really fantastic performer. Price f16, tax paid and carr. paid or with Stereo Decoder, $: 824$.

# GOING BACK... <br>  <br> COLIN RICHES ARTHUR DOW 

Abook which was recently submitted to us for review was entitled "A History of The Marconi Company" and published by Methuen \& Co. Ltd. The reason we felt we should mention it in the Going Back article and not in our normal New Books section is that not only is this book a history of a Company-it is an excellent reference to the history of radio.

The first chapter entitled 'The Stage is Set' describes the conception of the use of electromagnetic


Marconi and his Coherer Receiver-1896.


Marconi and Kemp with 1901 apparatus. (Copyright Marconi Company.) waves for communication purposes and covers the work of pioneers like James Clerk Maxwell (1855), our old friend Hertz and Professor (later to be Sir) Oliver Lodge, who in 1894 demonstrated his improved 'Branly' tube of iron filings in a new and important role-that of a detector of Hertzian waves. Lodge incidentally christened this device a 'coherer' because the filings 'cohered' whenever a Hertzian wave was applied to it by the associated circuitry.

Chapter two, 'The Young Signor Marconi' tells how Guglielmo Marconi, born on 25th April, 1874 at No. 7 Via delle Asse in Bologna, came to take interest in Professor Righi's papers on electromagnetic radiation. It describes how the critical point in Marconi's life came when he was on holiday in the Italian Alps and he chanced to read yet another paper on the Hertz experiments. Out of this, the idea to use the waves as a means of communication was born and this fired his enthusiasm to the point where he cut short his holiday and returned to his laboratory-a spare room at the Villa Griffone.
Marconi's first effort at commercial exploitation of his equipment was flatly turned down by the Italian Government and the book tells how, after a family conference, it was decided to send the twenty-one-year-old Marconi to England. This, they thought, would be a logical choice of country, for there were influential relatives living in London-the notable amongst these being Jameson Davis, who had promised all possible aid. Furthermore it was wellknown that Great Britain was the hub of a great Empire and the workshop of the world possessing the world's biggest mercantile fleet and the mightiest Navy-and it was in shipping that Marconi saw his best chance.
This was with good reason for, as the text states, it is difficult today to visualise the utter isolation which enshrouded a ship of the 1890's once it had lost sight of land. Disaster could strike-and not infrequently did-with no-one on shore or in nearby vessels being any the wiser.
The entry of Marconi into England was marred by the fact that an over-zealous Customs officer investigated his apparatus with such thoroughness that it arrived in London broken and useless. Repairs were soon made however and the equipment was demonstrated to Jameson Davis who set about establishing influential contacts.
On June 2, 1896 Marconi applied for the world's first patent for wireless telegraphy (Brit. Pat. No.

12,039). By this time, contact had been made with A. A. Campbell Swinton, an eminent electrical engineer of the day, who in turn provided Marconi with a letter of introduction to William H. Preece (later Sir William) the then Chief Engineer of the British Post Office.

Further chapters go on to describe how work progressed on Marconi's transmitters and receivers and the great trouble he had trying to cut red tape and make governments see the possibilities of wireless communication. The help that ships gained from Marconi's experiments was phenomenal and lighthouses became far less cut-off from the rest of the world when equipped with Marconi apparatus, but it seemed that governments remained sceptical.

The foundation of the Marconi Company is described. The first suggestion put forward for the name of the company was "Marconi's Patent Telegraphs Ltd.", this was rejected by Marconi who agreed it should be called "The Wireless Telegraph and Signal Company Ltd."

The remainder of the history of the Marconi Company is chronicled and the uses to which wireless apparatus were put are described in detail with magnificent pictures and diagrams throughout. There is even a photograph of the W.A.A.F. servicing a World War 2 bomber with the Marconi T1154/R1155 transmitter/receiver clearly shown. Just a point of in-terest-over 80,000 of these units were manufactured!

And so we read on through the book and come


## The letter of introduction to William H. Preece.



The wireless cabin of the Lusitania (copyright Marconi Company).


The Poldhu aerial system erected for the transatlantic experiment of 1901.


The first 2LO at Marconi House, used for transmissions from 1922. On the left is the rectifier. Next is the master oscillator with "cage-mounted" valve. Third panel contains main oscillator with h.f. circuit and speech choke. Main and sub-modulator with five valves are to the right.
across such chapters as 'The Passing of Guglielmo Marconi', 'Developments in Radar', 'Developments in TV', 'Developments in Aviation and Marine Communication'. Therefore, all in all, we feel that this book will not only be of interest-and great interest at that-to the vintage radio enthusiast, but almost a Bible to all those who take any interest in wireless communication, however minute.

For those readers who are interested, the book, "A History of the Marconi Company" by W. J. Baker is published by Methuen and costs $£ 5$.

## $\mathfrak{C} \mathbb{A}!\mathbb{C} \mathbb{R}!\mathbb{C} \mathbb{R}!\mathbb{C} \mathbb{Q}!\mathbb{C} \mathbb{R}!$

## BOOKS WANTED

. . . a copy of "Outline of Wireless" by Ralph Stranger. I have often wondered who Ralph Stranger really was or-is. He is one of the best writers I have ever come across.Ge is one of Watson, 2 Winn Court, Winn Road, Southampton, SO2 1UZ.

## INFORMATION REQUIRED

any information on the Dorian Super Screened Four. Valve line-up is: HL2, LP2, 210HL, PM12M. I would also welcome advice and possible source of supply of a jelly acid non-spillable accumulator-for, as I wish to keep the set as an original example, I do not propose making modifications. -D. Claridge, 3 Bakery Drive, Horsefair, Shrewsbury, Shropshire.

# TRANSISTOR RADIOS 

 TO BUILD YOURSELF
## Backed by after sales service

## NEW! roamer eight mk 1 WITH VARIABLE TONE CONTROL

7 Tunable Wavebands: Međinm Wave 1, Medium Wave 2, Long Wave, swl, SW2, sw3 and Trawler Band. Built in Ferrite Rod Aerial for Medium and Long Waves. 4 section $24^{\prime \prime}$ retractable chrome plated Teleacopic serial for Short Waves for maximum performance. Push pull output using 600 mW trandistors. Socket for car aerisi. Tape record socket. Selectivity switch. Switched earpiece socket complete with earpiece for private listencondenser. On/Oft awitch volume control. Wave change switch and tuning control. Attractivg condenser. tions and diagrams make the Roamer Eight a pleasure to Total Building Costs Parts Price List and Easy Build Plans 5/- (EREE with parts).
overseas P. \& P. 17/

## roamer seven mk IV

SEVEN FULLY TUNABLE WAVE. BANDS-MW1, MW2, JW, SW1, SW2, SW3 and Trawler Band. Extra Medium waveband provides easier tunitig of Radio Juxembourg, etc. Built in ferrite rod aerial for Medjum and Long Waves Retractable 4 sectron 24in, chrome plated telescopic Socket for Car Aerlal. Powertul puahpull output. Seven trangistors and two diodes facluding Micro-Alloy R.F. Transistors. Famous make $7 \times 4 i n$. P.M. spealer. Air spaced ganged tuning condenser. Volume/on/of control, wave change switches and tuning control Attractive case with carrying handle. Size $\theta \times 7 \times 4 \mathrm{in}$.
approx. Easy to follow instructions and dingrams make the
Roamer 7 a pleasure to build. Parts price list and easy build plans 3/- (FREE with parts): Personal Earpiece with switched socket for
private listening, $\delta /-$ extra.


## Total building costs <br> £5.19.6 <br> P. \& P. 7/6 <br> Overseas P. \& P. 17/-

## pocket five

MEDIUM WAVE, LONG WAVE AND TRAWLER BAND PORTABLE WITH SPEAKER
Attractive black and gold case. Size $\left.5 \frac{k_{2}}{} \times 1\right\} \times 31 \mathrm{in}$. Tunable over both Melium and Long Waves with extended M.W. band for easier tuning of Luxem-

sumersensitive fertite rod aerial, fine tone dioving coil speaker. Easy build plans and parts price list 1/6 (FREF; with parts).

Total building costs
AAA\& $\begin{gathered}P . \& P \\ 3 / 6\end{gathered}$
Overseas P. \& P. 10/-

## IMPROVED MODEL!

 roamer sixSIX WAVEBAND PORTABLE WITH 3in. SPEAKER
Attractive black case with red grille and black knobs and dial with spun brass inserts. Size $9+34+2 \frac{3}{8}$ in. approx. Tunable on Mediun and Long Waves, two
Short Waves, Trawler Band plua an extra M.W. band for easier tuning of Luxembourg, etc. Sensitive ferrite rod serial and jatest telescopic aarial for
Short Waves. Improved circuit. 8 stages- 6 tranBhort Waves. Improved circuit. 8 stages- 6 transistors and 2 diodes including Micro-Alloy R.F. Transiators, etc. Easy build plans and parts price
list $2 /-$ (FREE with parts).

* Callers side entrance Stylo Shoe Shop
$\star$ Open IO-I, 2.30-4.30 Mon.-Fri. 9-I2 Sat.


## NEW!

trans eight SIX WAVEBAND PORTABLE WITH 3in. SPEAKER
Attractive case in black with red grille and black knobs inserts. Bize $9 \times 5 t \times 2$ in appror
Tunable on Medium and Long Waves, three
Short Waves and Trawler Band. Sensitive ferrite red
aerial for M.W. and L.W. Telescopic aerial for Short Waves.
Eight improved type tradeistora plua 3 diodes. Push pull output.
Aattery econoniser switch for extended battery life. Ample power to drive a larger speaker. Parts price list and easy build plans 5/- (FizEE with parts). Earpiece with switched socket for private listening 5/- extra.
$\left.\begin{array}{c}\text { Total building costs } \\ \text { P. \& P. } 5 / 6\end{array}\right\}$
Overseas P. \& P. 13/-

## transona five

MEDIUM WAVE, LONG WAVE AND TRAWLER BAND PORTABLE
WITH SPEAKER

Attractive case with red speaker grille. Size $6 \frac{1}{2} \times 4 \frac{1}{2} \times 1$ itin. 7 stage- 5 transistors and 2 diodes, ferrite rod aeria, tuning condenser volume control, fine tone moving coll speaker Easy build plans and parts price list $1 / 6$
(FREE with parts).

## RADIO EXCHANGE CO

|61 HIGH STREET, BEDFORD. Tel. 023452367


| ROAMER EIGHT | $\square$ | ROAMER SEVEN |
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| TRANSONA FIVE | $\square$ | TRANS EIGHT |
| POCKET FIVE | $\square$ | ROAMER SIX |

\|Parts price list and plans for


## Z. 30 \& Z. 50 power amplifiers

The $Z .30$ together with the $Z .50$ are both of advanced design using silicon epitaxial planar transistors to achieve unsurpassed standards of performance. Total harmonic distortion is an incredibly low $0.02 \%$ at full output and all lower outputs. Whether you use the 2.30 or Z.50 power amplifiers in your Project 60 system will depend on personal preference. but they are the same physical size and may be used with other units in the Project 60 range equally well. For operating from mains. for the $Z .30$ use PZ. 5 for most domestic requirements, or PZ. 6 if you have very low efficiency loudspeakers. For Z.50, use the PZ 8 described below.

SPECIFICATIONS (Z. 50 units are interchangeable with Z.30s in all applications). Power Outputs
Power
Z. 3015 watts R.M.S. into 8 ohms. using 35 V 20 watts R.M.S. into 3 ohms using 30 volts.
75040 watts RMS into 3 onms from 40 volts Z. 5040 watts R.M.S. into 3 ohms from 40
30 watts R.M.S.into 8 ohms, using 50 volts.

Frequency response 30 to $300.000 \mathrm{~Hz} \pm 7 \mathrm{~dB}$ Distortion $0.02 \%$ into 8 ohms
Signal to noise ratio better than 70 dB unweighted Input sensitivity 250 mV into 100 Kohms .
For speakers from 3 to 15 ohms impedance. Size $3 \frac{1}{2} \times 2 \frac{1}{4} \times \frac{1}{2}$ ins.

2.30

Built, tested and guaranteed with crrcuits and instructions manual 89/6
2.50

Bult. tested and guaranteed with circuits and instructionsmanual $100 / 6$

## Stereo 60 pre amp/control unit

Designed for the Project 60-range but suitable for use with any bigh quality power amplifier. Again silicon epitaxial planar transistors are used throughout, achieving a really high signal-to-noise ratio and excellent tracking between channels. Input selection is by means of push buttons and accurate equalisation is provided for all the usual inputs.

## SPECIFICATIONS

- Input sensitivities - Radio - up to 3 mV . Mag. pu. 3 mV : correct to R.I.A.A. curve $\pm 1 \mathrm{~dB}: 20$ to $25,000 \mathrm{~Hz}$. Ceramic p.u. - up to $3 m$ V: Aux. - up to 3 mV .
- Output-250mV.
- Signal-to-noise ratio-better than 70 dB .
- Channel matching - within 1 dB .
- Tone controls -- TREBLE +15 to -15 dB at Tone controls - TREBLE +15 to
10 kHz : BASS +15 to -15 dB at 100 Hz .


## Active Filter Unit

For use between Stereo 60 unit and two Z.30s or Z.50s, the Active Filter Unit matches the Stereo 60 in styling and is as easily mounted. It is unique in that styling and is as easily mounted. It is unique in that the cut-off frequencies are continuousiy variable. ( $12 \mathrm{~dB} /$ octave). there is less loss of the wanted signal than has previously been possible. Amplitude and phase distortion are negligible. The Sinclar and phase distortion are negligible. The Sinclar
A.F.U. is sutable also for use with any other amplifier system.
Two stages of filtering are incorporated - rumble
(high pass) and scratch (low pass). Supply voltage75 to 35 V . Current -3 mA . H.F cut-off ( -3 dB )


- Front panel - brushed aluminium with black knobs and controls.
- Size $8 \frac{1}{4} \times 1 \frac{1}{2} \times 4$ ins.

Buit, tested and guaranteed
£9.19.6

## Stereo FM tuner



## first in the world to use the phase lock loop principle

Before production of this tuner, the phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio over other systems. Now, for the first time the principle has been applied to an FM tuner with fantastically good results. By the inclusion of other original features such as varicap diode tuning. printed circuit coils and an I.C. in the specially designed stereo decoder, the tuner has an unsurpassed specification, which also incorporates a squelch circuit for silent tuning between stations, A.F.C. and A.G.C. Sensitivity is such that good reception becomes possible in difficult areas, foreign stations can be tuned in suitable conditions and often a few inches of wire are enough for an aerial. In terms of high fidelity. this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically as the tuning control is rotated, a panel indicator lighting up as the stereo signal is tuned in. Although the tuner is intended primarily for use with a Project 60 system, it can be used to advantage with any other high fidelity system. It is easily mounted into any cabinet as shown in the manual supplied with it.

## Specifications

Number of transistors 16 plus 20 in I.C.
Tuning range 87.5 to 108 MHz
Capture ratio 1.5 dB
Sensitivity $2 \mu \vee$ for 30 dB quieting
$7 \mu \vee$ for fulllimiting
Squelch level $20 \mu \mathrm{~V}$
A.F.C. range $\pm 200 \mathrm{KHz}$

Signal to noise ratio $>65 \mathrm{~dB}$
Audio frequency response $10 \mathrm{~Hz}-15 \mathrm{kHz}( \pm 1 \mathrm{~dB})$
Total harmonic distortion $0.15 \%$ for $30 \%$
modulation
Stereo decoder operating level $2 \mu \mathrm{~V}$
Pilot tone suppression 30 dB
Cross talk 40 dB
I.F. frequency 10.7 MHz

Output voltage $2 \times 150 \mathrm{mV}$ R.M.S.
Aerial Impedance 75 Ohms
Indicators Mains on : Stereo on : tuning indicator Operating voltage 25-30 VDC
Size $3.6 \times 1.6 \times 8.15$ inches: $91.5 \times 40 \times 207 \mathrm{~mm}$


Price : $\mathbf{2 5}$ built and tested. Post free.

## Power Supply Units

The units below are designed specially for use with the Project 60 system of your choice.
Illustration shows PZ. 5 power supply unit to left and $P Z .8$ (for use with $Z .50$ s) to the right. Use PZ. 5 for normal Z.30 assemblies and PZ. 6 where a stabilised supply is essential.

PZ-5 30 volts unstabilised $£ 4.19 .6$
PZ-6 35 volts stabilised $£ 7.19 .6$
PZ-8 45 volts stabilised
(less mains transformers) £5.19.6
PZ-8 mains transformer $£ 5.196$
GUARANTEE if within 3 months of purchasing Project 60 moduies directly from us, you are dissatisfied with them. we will refund your money ar once. Each module is guaranteed to work perfectly and should any defect arise in normal use we will service $h$ at once and without any cost to you whatsoever provided that it is returned to us within 2 years of the purchase date. There will be a small charge for service thereafter. No charge for postage by surface mail. Air-mail charged at cost.
ㄷirnclair

# Project 60 



## the world's most advanced high fidelity modules

With the introduction of an entirely new and original high fidelity stereo F.M. tuner, the Project 60 range can be said at this stage to be complete. It offers the constructor a most attractive choice of modular arrangements whereby a high fidelity system can be selected to suit the user's personal requirements. Equally, it is possible to use any Project 60 modules separately or partially grouped and so benefit greatly from the flexibility in use these modules afford. The chart below shows some of the most popular applications for constructors to assemble. The Project 60 manual (free with the modules) suggests others as well and its 48 pages are packed with valuable information. The new tuner. for example can be used with any good high fidelity system as well as Project 60.

Project 60 now falls into four interdepencent groups: -1 . The $Z .30$ and $Z .50$ amplifiers which have only $0.02 \%$ distortion at all output levels and are useful in a wide variety of other applications. 2. The control units comprising the Stereo 60 preamp and control unit and the Active Filter Unit (A.F.U.) with which both high pass and low pass filtering can be introduced between control unit and power amplifiers. 3. The Stereo F.M. tuner as described opposite : and 4. The power supply units PZ.5.
$P Z .6$ and $P Z .8$. For most requirements when using $Z .30$ power amplifiers, the PZ. 5 will be perfectly adequate ; if low efficiency (high quality) loud speakers are used, the PZ. 6 stabilised power supply unit will be used. The PZ. 8 will be needed with Z.50s which can be used for any Project 60 system.
Project 60 modules incorporate some of the most advanced circuitry in the world to achieve unsurpassed standards of high fidelity and modern manufacturing techniques enable these modules to be sold at exceptionally attractive prices. Assembling the modules requires no skill or previous experience since the manual supplied with the modules explains clearly how everything can be done with nothing more than the simplest of domestic tools.

## Project 60 manuais

How to assemble and use Project 60 modules to best advantage in the above and other applications will be found in the fully descriptive Project 60 manual included with Project 60 systems. This 48 page manual is available separately, price $2 / 6 d$ including postage.

|  | System | The Units to use | In conjuction with | Cost of Units | + Project 60 tuner |
| :---: | :---: | :---: | :---: | :---: | :---: |
| A | Car Radio | Z.30 | Existing car radio. Sinclair Micromatic | 89/6 |  |
| B | Simple battery powered record player | Z.30 | Crystal pick-up, 12 V or more battery supply and volume control | 89/6 |  |
| C | Mains powered record player | Z.30 and PZ. 5 | Crystal or ceramic P.U. <br> Volume control etc. | £9.9.0 | £34.9.0 |
| D | $20+20$ watts R.M.S. stereo amplifier for most needs | Two Z.30s, Stereo 60 and PZ. 5 | Crystal, ceramic or magnetic P.U., most dynamic speakers, F.M. tuner etc. | £23.18.0 | £48.18.0 |
| E | $20+20$ watts R.M.S. stereo amplifier for use with low efficiency (high performance) speakers | Two Z.30s, Stereo 60 and PZ. 6 | High quality ceramic or magnetic P.U.. F.M. Tuner, Tape Deck, etc All dynamic speakers | £26.18.0 | £51.18.9 |
| F | $40+40$ watts R.M.S. de-luxe stereo amplifier | Two Z.50s, Stereo 60 PZ. 8 and mains transformer | As for $E$ | £32.17.6 | £57.17.6 |
| G | Outdoor public address system | Z.50 | Microphone. up to 4 P.A. speakers, 12 V car battery with converter, or 45 V d.c., controls | £5.9.6 |  |
| H | Indoor P.A. | One Z.50, PZ.8 and mains transformer | Microphone, guitar, heavy duty speakers etc., controls | f17.8.6. |  |
| $J$ | High pass and low pass filters | A.F.U. | D. E or F as above | £5.19.6 |  |

## Sinclair IC10/Q16/Micromatic

## IC10



The world's most advanced high fidelity amplifier
The circuit comprises a minute chip of silicon giving 5 watts RMS output (10 watts peak) which contains 13 transistors (including two power types), 2 diodes, 1 zener diode and 18 resistors. This exciting device is not only rugged and reliable, it also has performance advantages including freedom from thermal runaway and very low level of distortion. The IC10 is primarily intended as a full performance high fidelity power and pre-amplifier, for which it only requires the addition of such components as tone and volume controls and a battery or mains power supply. It may also be used simply in many other applications inciuding car radios, electronic organs, servo amplifiers (it is dc coupled throughout), etc.

## Circuit Description

Three transistors form the pre-amp and the remaining 10 the power amplifier. Generous neg. feedback is used round both sections and the amplifier is free from crossover distortion at all supply voitages, making battery operationeminently satisfactory.
Applications
Each IC10 is sold with a comprehensive manual giving circuit and wiring diagrams for a large number of applications in addition to high fidelity. These include oscillators. RF or IF amplifier without any additional transistors.
Specifications
Output: 10 watts peak, 5 watts RMS continuous.
Frequency response: 5 Hz to $100 \mathrm{kHz} \pm$ 1 dB .
Total harmonic distortion: Less than $1 \%$ at full output.
Load impedance: 3 to 15 ohms.
Power gain: 110 dB ( $100,000,000,000$ times) total.
Supply voltage: 8 to 18 volts.
Size: $1 \times 0.4 \times 0.2 \mathrm{In}$. plus heat sink and tags. Sensitivity: 5 mV .
Input impedance: Adjustable externally up to 2.5 Mohms.
Price, with manual, $59 / 6$ post free.


## High fidelity loudspeaker

The Q16 employs well proven acoustic principles in which a special driver assembly is meticulously matched to the physical characteristics of the uniquely designed housing. Technical journals have been loud in their praise for this speaker which comfortably stands comparison with very much more expensive models. Its shape enables it to be positioned and matched to its environment to much better effect than is the case with conventionally styled enclosures, and with its improved styling, the Q16 presents an entirely new and attractive appearance. A solid teak surround is used with a special all-over cellular foam front chosen as much for its appearance as for its ability to pass all audio frequencies.
The Q16 makes the ideal shelfmounted speaker, and brings genuine high fidelity within reach of every music lover.

Driver unit: Specially designed high compliance unit with aluminium speech coil and special cone suspension. Size and styling: $9 \frac{3}{4} \mathrm{in}$. square on face $x 4^{\frac{3}{2}}$ in. deep with neat pedestal base; all-over cellular foam front with natural solid teak surround.
Frequency response: From 60 to 16,000 Hz , as confirmed by independently plotted $B$ and $K$ curve.

## Specifications

Loading: Up to 14 watts RMS.
Input impedance: 8 ohms.
Price: £8.19.6.

## Micromatic



Britain's smallest radio
Considerably smaller than an ordinary box of matches, this is a fantastically efficient multi-stage AM receiver meticulously designed to provide remarkable standards of selectivity, power and quality. Powerful $A G C$ is incorporated to counteract fading from distant stations, whilst bandspread at higher frequencies makes reception of Radio 1 easy at all times. The high quality plug-in magnetic earpiece provided matches the Micromatic's output to provide wonderful standards of reproduction. Everything including the special ferrite rod aerial and batteries is contained within the very small and attractively designed case. Whether you build a Micromatic kit or buy this amazing receiver ready built, tested and ready for immediate use you will find it as easy to take with you as your wrist-watch, and dependable under the severest operating conditions.
Specifications
Size: $46 \times 33 \times 13 \mathrm{~mm}$. ( $\left.1 \frac{14}{6} \times 1 \frac{3}{10} \times \frac{1}{2} \mathrm{in}.\right)$. Weight: including batteries, 28.4 gm (one ounce).
Case: Black plastic with aluminium front panel and dial.
Tuning: Medium wave band.
Earpiece: Magnetic type.
Kit in pack with earpiece, case, instructions and solder 49/6.
Ready built, tested and guaranteed, with earpiece 59/6.
Two Mallory Mercury batteries type RM675 required. Available from radio shops, chemists, etc.

Sinclair Radionics Ltd, London Rd, St lves, Huntingdonshire. Tel: St lves (048 06) 4311
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ADDRESS E Please send
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for which $f$ enclose cash/cheque/money order.

## electric clock

WITH 25 AMP SWITCH Made by Smith's, tbese units ara as fitted to many top quality clock is mains driven and frequency controlled so it is ex-
trenuely aceurate. The two small tremely aceurate. The two small
dials enable switch on and off dials enable switch on and off times to be accurately set. Ideal sasisasy for switching on tape recorders. Offered at only a
fraction of the regular price-new and unused only 39/6, Iess than the value of the clook alonepost and hinsurance $2 / 9$.

## FLUORESCENT CONTROL KITS

Each kit comprises seven items-Choke, 2 tube ends, starter, starter holder and a tube
clips, with wiring instructions. Suitable for clips, with wiring instructions. Suitable for
normal fluorescent tubes or the new "Grolux" normal fluorescent tubes or the new
tubes for fish tanks and indoor plants. Choke tubes for fish tanks and indoor plants. Chozes $-15-20 \mathrm{w} .19 / 6$. Kit B-30-40 w. 19/6. Kit C-80w. 28/6. Kit E-65w. 28/6. Kit F for 8 ft .125 w . tube $35 /$-. Kit MFl is for
6 in . 9 in . and 12 in . niniature tubes, $19 / 6$. Kit Min. and 12 in . miniature tubes, $19 / 6$ Kostage for Kits A 13 w. miniature tube 20/Postage on Kits $A$ and B $4 / 6$ for one or two kits then $4 / 6$ for each two kits ordered. each hit ardered. Kit $F 6 / 6$ than $4 / 6$ tor each each kit ordered. Krder Kit MFI $3 / 6$ on first kit then $3 / 6$ on each two kits ordered.

BLANKET SWWTCH
Double pole with neon let
into side so luminous in daris, ideal for dark room light or for plastic case. $5 / 6$ each. 3 heat model 7/6.

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Although looking like, and fitted as an ordinary blanket switeh, this is in fact a device for awitching on for varying time periods, thus giving a complete control from off to full heat. Although suitable for controlling the temperature of any other appliances using up to 1 amp. Listed at last at only $12 / 8$ each.
REED SWITCHES Class encased, switclies operated by external 3 types:
Miniature. Tin. long $\times$ approximately tin. Miniature. Will make and break up to $\frac{1}{2} A$ up to 300 volts. Price $2 / 6$ each. 24/- dozen. Standerd. 2in long $\times 3 / 16 \mathrm{in}$. diameter. This will break currents of up to 1 A , voltages up to 250 Volts. Price 2/- each. 18/- per dozen. $1 / 16$. thick Flat. Mat type, 2 in . long, just over $\operatorname{approximately}$ fin. wide. The Standard Type appartened out, so that it can be fitted into a smaller space or a larger quantity may be packed into a square solenoid. Rating 1 amp 200 volts. Price 6/- each. \& per dozen. Small ceramic magnetis to operate these reed switches $1 / 9$ each $18 /=$ dozen

HIGH CAPACITY ELECTROLYTICS Trand. new, not ex-equipment.
$100 \mathrm{mfd} .25 \mathrm{v} ., 1 / 8$ each $12 /-\mathrm{doz}$ $100 \mathrm{mfd} .25 \mathrm{v} ., \frac{1 / 8}{}$ each $12 /-$ doz. 200 mif . $25 \mathrm{v} ., 2 / 8$ each $250 \mathrm{mfd} .50 \mathrm{v} ., 3 / 8$ each $88 /-\mathrm{doz}$.
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$6000 \mathrm{mfd} .12 \mathrm{v}, 4 / 9$ each $\mathbf{~} 2.8 .0 \mathrm{doz}$. $10,000 \mathrm{mfd}$. 6 v ., $5 / 9$ each $\$ 8.0,0$ doz. $10,000 \mathrm{mfd}$. 15 v., $8 / 6$ each $\$ 4.10 .0$ doz. $15,000 \mathrm{mfd}$. $10 \mathrm{v} ., 10 / 6$ each $\$ 5.0 .0 \mathrm{doz}$. $60,000 \mathrm{mfd}$. 8 V ., $22 /-$ each 510.0 .6 doz. $70,000 \mathrm{mfd} .18 \mathrm{~V} ., 40 /-$ each $\$ 20.0 .0$ doz 3 amp 12v Eattery Charger Kit-comprising 230/40 mains transformer with 3 amp secondary and 3 amp rectifier $2 / 6$ plus $4 / 6$ post.
12 volt $1 \frac{1}{2} \mathrm{v}$ amp Power Pack. This compriges double-wound $230 / 240 \mathrm{~V}$ mains trensformer with full wave rectifler and $2000 \mathrm{~m} / \mathrm{f} / \mathrm{d} / \mathrm{smoothing}$. Price 27/6.
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you never know when you will need some. Famous make, brown bakelite, standard size, 12 for 13/plas 4/6 post.
Ditto bat with switch. 12 for $\$ 1$ pius $4 / 6$ post. 13 amp sockets, flush mounting. Bakelite, cream, less switch. 6 for 81 .
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$230 / 240$ volt mains operated Clock and a 20 amp Srrit 230/240 volt mains operated Clocs and a 20 amp Swritch, the switeh-off time of Which can be delayed up to 12 hours (continuonsly variable not stepped) $3 \frac{1}{4} \times 2$ in , aeep. Metal encased, glass fronted with chrome surround. Offered at $\times 22 \mathrm{in}$, deep. Metal encased, glass

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Is undoubtedly one of the finest loudspeasers that we have ever offered, produced by one of the conntry's most famous makers. It has a die-cast metal frame and is strongly recommended for Hi-Fi and public address. Handling 40 watts R.M.S.-Cone moulded Albre-Freq. response 30 $10,000 \mathrm{c} . \mathrm{p} . \mathrm{s}$.-spectify 3 or 15 ohms. Chassis diam. 12 in .oftered this month for $\mathbf{f}^{6,19.8}$ plus 7/6 post and ins.


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Just what you need for work bench or lab. $4 \times 13 \mathrm{amp}$ sockets in metal box to take standard 13 amp fused plags and on/off switch with neon warning light. Suppled complete with 7 feet of heary cable. Wired up ready to work, $89 / 6$ less plug complete with $15 /-$ fith fitted 13 amp plug; $47 / 6$ with fitted 15 amp plug, plus $4 / 6$ Z. \& I.

BARGAIN OF THE YEAR

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 7 transistor Keychain Radio in very pretty case, size $2 \frac{\pi}{4} \times 2 \frac{1}{2} \times 1 \frac{1}{4} \mathrm{in}$. complete with leather zipped bag. Specification:Clrcuit: 7 transistor superheterodyne.Frequency range: 530 to $1000 \mathrm{Kc} / \mathrm{s}$. SensiFrequency range: 530 to $1600 \mathrm{Kc} / \mathrm{s}$. Sensi-
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$465 \mathrm{Ke} / \mathrm{s}$ or $455 \mathrm{Kc} / \mathrm{s}$. Power output: 40 mW . Antenna: ferrite rod. Loudspeaker Permanent magaet type. In transit from the East these sets suffered alight corrosion as the batteries were left in them but when this corrosion is cleared away they
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These heater unts are the very latest type, most efficient, and quiet running. As fitted in Hoover prises motor, impeller and two alenents allowing variable heat switching and with thermal saiety cut out. Can be fitted into any metalline case or cabinet. Only need control awitch. $1 \frac{3}{4} \mathrm{~K} . \mathrm{W}$. model
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 Neat flat toroh. fits unobtrosively in your pocket,contains 2 Nicad cells and built-in charger, Plugs into shaver adsptor and charger from our standard $200 / 240$ volt mains. American made, sold originally at over 4 dollars. Our price only 19/6 each.


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 Corner Cabinet $20 \frac{3}{4} \times 13 \times 7 \frac{3}{2}$ itı. deep. Vynalr front. Fitted rubber feet. Five speaker units 15 ohms, impedence handles
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ADASTRA DOUBLE 5 stereo solid state amplifier housed in handsome cabinet veneered in natural teak. Size $11 \frac{1}{2} \times 6 \times 5 \frac{1}{2}$ in. 10 Transistors-power outpur watts (our Cowdrey speaker system eminently suitable). Smart biue escutcheon. f14.14.0. P. \& P. 10/6. SCOTT. This elegant tapered cabinet $10 \frac{1}{2} \times 16 \times 5 \frac{1}{2}$. deep is attractively finished In black cloth with striped grey Vynair front. Sultable for table or for wall mounting. Fitted with $13 \frac{1}{2} \times$ Bin. speaker unit and volume control, 3 or 15 ohms impedence-please state impedance required. £4.15.0. P, \& P. 76 d өach. Fited with E.M.. $15 \times \mathrm{hm}$. lmpedence , capacity 10 watts. 46/= extra.


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THE FIESTA 59/6 Brit. Isles.

Fitted with Vol ume control. Teak veneered with Black Vynair ume control. Teak veneered with Black $10^{\prime \prime} \times 7 \frac{1}{2} \times 3 \frac{1}{2}$ " deep. 3 ohm ex TV Speaker.


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Size $21 \times 11 \times 6 \frac{1}{2}$ in.
An extremely elegan speaker system made covered with teak eathercloth with deark or grey Vynalir front send S.A.E. for cloth samples). This three ex $T V$ speaker Carefully matched and tested. Will and will match 8 ohms impedance. If
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Transistorised Stereo HI-Fi Record Player
Build your own Hi-Fi Record Player with the Serenade fully tran. sistorised amplifier which comes complete with $2-10$ in. x 6 in. speakers and the latest BSR Speed StereolMono Record Changer. 9 : Gns. Credit terms ist monthly payment 63.6 .2 d . followed by $4017 / 6$ Send $\pm 4.3 \mathrm{~s} .8 \mathrm{~d}$. today.
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Phono piugs, red, white, yellow, green, grey, blue, black 11d $\theta /-$ per doz; Coarial plugs, aluminium $1 / 6$; Consial couplers $1 /-; 2$ pin DIN plugs $8 /-; 3$ pin DIN plugs $8 /-; 5$ pin DIN plugs $3 / 5 ; 2$ pin, 3 pin DIN sockets $1 / 6$; wander Plugs 88; sockets 8d; Banana plugs 1/1, sockets 1/-; 3.5mm J/Pluge 1/6 \& 2/6; Btandard J/Plugs with solder terminals 2/-; Chrome 3/-; Side Entry J/Plugs black and chrome 4/月; Insulated j/sockets, o/ect 3/3, c/ect 8/6.

GLEATROLYTIC CAPACITORS
4uF, 150v 9d; 8uF 500v 8/-; 8uF 12v 10d; BuF + $8 \mathrm{uF} 450 \mathrm{v} / 3 ; 10 \mathrm{uF} 150 \mathrm{v}$ 8d; 12 uF 25 v 9d; 16 uF $450 \mathrm{v} 2 / 9 \mathrm{~d} ; 16 \mathrm{mF}+16 \mathrm{uF} 450 \mathrm{v} 4 / 8 ; 30 \mathrm{uF} 10 \mathrm{v} 8 \mathrm{~d} ; 50 \mathrm{uF}$ 10 v 9 d ; $100 \mathrm{uF} 9 \mathrm{v} 9 \mathrm{~d} ; 100 \mathrm{uF} 12 \mathrm{v} 9 \mathrm{~d} ; 150 \mathrm{uF} 12 \mathrm{v} 9 \mathrm{~d}$; $320 \mathrm{uF} 9 \mathrm{v} \mathrm{1/-;} \mathrm{1500uF} \mathrm{30v} \mathrm{3/9;} \mathrm{2.5uF} \mathrm{16v} \mathrm{6d;}$ 10uF 16v 1/6; $5000 \mathrm{uF}^{25 v} 10 /-$; $10,000 \mathrm{uF} 25 \mathrm{v}$ 15/-; 20,000uF 30 v 21/-.

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Acos Mic 45 22/6; Acos Mie 60 19/11; Planet CM70 30/-; Hand Mike 15/-; Shure 201 s.5; Shure 444 \$10.10.0.

Valves
PY81 7/2; PCL82 7/9; EF183 9/-; EF184 9/-.
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| 2N3055 | 15\%- | 2 N 5459 | 9/9 | BA130 | $4 / 6$ | BF195 | 7/6 | ZTX 501 | 5/- |
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| - 2 N 3663 | 11/6 | 40361 40362 | $12 / 8$ | ${ }_{\text {BC108 }}$ | 2/6 | - | 8/9 | ZTX503 | 5/- |
| 2N3703 | $2 / 6$ | 40406 | 16/3 | BC109 | $2 / 9$ | BFX87 | 5/8 | ZTX 504 | 12/- |
| 2N3704 | 2/6 | 40408 | 14/6 | 3C125 | 11/- | BFX88 | 8/3 | ZTX 530 | 5/5 |
| 2N3705 | 2/6 | 40430 | $3{ }^{12} /-$ | $\mathrm{BCl}^{26}$ | 11/- | BFY50 | 4/6 | ZTX531 | 8/9 |

## RESISTORS

| Code | Power | Tolerance | Range |
| :---: | :---: | :---: | :---: |
| c | 1/20W | $5 \%$ | $82 \Omega-220 \mathrm{~K} \Omega$ |
| $\mathrm{C}^{\mathrm{C}}$ | 1/8W | 5\% | $4.7 \Omega-330 \mathrm{~K} \Omega$ |
| c | $1 / 4 W$ | 10\% | $4.7 \mathrm{R}-10 \mathrm{M} \Omega$ |
| g | 1/2W | 5\% | $4.7 \Omega-10 \mathrm{M} \Omega$ |
| C | IW | 10\% | $4.7 \Omega-10 \mathrm{M} \Omega$ |
| M0 | 1/2W | 2\% | $10 \Omega-1 \mathrm{M} \Omega$ |
| WW | 1W | 10\% $\pm 1 / 20 \Omega$ | $0 \cdot 22 \Omega-3.9 \Omega$ |
| WW | 3 W | 5\% | $12 \Omega-10 \mathrm{~K} \Omega$ |
| ww | 7W | 5\% | $12 \Omega-10 \mathrm{~K} \Omega$ |

$\begin{aligned} \mathrm{C} & =\text { carbon film high stability low noise } \\ \text { MO } & \text { = metal oxide Electrosil TR } 5 \text { ultra low noise } \\ \text { wW } & =\text { wire wound Plessey. }\end{aligned}$
Velues:
E12 denotes reries: $10,12,15,18,22,27,33,39,47,56, ~$
63, 82 and their decades.
E24 denotes series: as E12 plus 11, 13, 16, 20, 24, 30, 36 , $43,51,62,75,91$ and their decades.

## INTEGRATED CIRCUIT

## AMPLIFIERS

SINCLAIR 1010 complete with instruction book giving amplifier circuit details and range of applications.

59/8 nett.
Components pack for stereo inc. mains transformer, controls etc. f4.15.0 nett
PLESSEY SL403A Now only $42 / 6$ nett 3W into $7.5 \Omega$ for 18 V supply. Application data sent with two more.

## WAVECHANGE SWITCHES

 LONG SPINDLES1P 12W: 2P 6W; 3P 4W; 4P 3W
SLIDER SWITCEIES D.P.D.T.
NEON INDICATOR LAMPS
all $200 / 250 \mathrm{~V}$. Square bezel, red only
Round chrome bezel red, amber, clear
4/0 each

19 each
Toggle switches. 250V a.c. 1.5A.
chrome dolly and chrome milled nut S.P.S.T
4/6; D.P.D.T. $5 / 9 ;$ S.P.D.T. centre off $5 /-$
S-DeO's put an end to "birdsnesting". Components just plug in Saves valuable tirne. Use components again and increased capacity, may be temperature-cycled. T-DeC (208 points) only $50 / \sim$. Full range stocked.

| Values available | 1 to 9 | 10 to 99 (see note below) | 100 up |
| :---: | :---: | :---: | :---: |
| E12 | 18 | 16 | 15 |
| E24 | $2 \cdot 5$ | 2 | 1.75 |
| E12 | $2 \cdot 5$ | 2 | 1.75 |
| E24 | 3 | $2 \cdot 5$ | $2 \cdot 25$ |
| E12 | 6 | 5 | 4.5 |
| E24 | 9 | 8 | 7 |
| E12 |  | 1/6 all quantities |  |
| E12 |  | 1/6 all quantities |  |
| E12 |  | 1/9 all quantities |  |

Prices are in pence each $1 / 9$ all quantities rating, NOT mixed values. (Ignore fraclions of 1 la. on tolal
resisfor order)

## MULLARD polyester C280 series

2d 20\%: 0.01; 0.022; 0.033, 0.047 8d ea. 0-068; 0.1 2d ea., 015 11d., $0.221 /-, 10 \%: 0.331 / 5,0.471 / 8,0.68$

MULLARD SUB-MIN ELECTROLYTIC G426 range arial lead
$1 / 3$ each
 4/40;5/64; $5 \cdot 4 / 6 \cdot 4 ; 6-4 / 25 ; 8 / 4 ; 8 / 40 ; 10 / 2 \cdot 6 ; 10 / 16 ; 10 / 64 ;$ 22-0/40; $32 / 64 ; 40 / 16 ; 40 / 2.5 ; 50 / 6 \cdot 4 ; 50 / 25 ; 50 / 40 ; 64 / 4$; $64 / 10 ; 80 / 2.5 ; 80 / 16 ; 80 / 25$; $100 / 6 \cdot 4$; $125 / 4 ; 135 / 10$; 125/16; 160/2-5; 200/6-4; 200/10; 250/4; 320/2.5; 320/6-4: 400/4; $500 / 2 \cdot 5$.

## LARGE CAPACITORS

High ripple current types: 1000/95 5/6; 1000/50 8/2: 2500/70 19/8: $5000 / 25$ 12/6: $5000 / 50$ 21/11: $5000 / 10058 / 3$; 10000/15 17/-; 10000/25 24/6; 10000/50 44/-; 10000/70 61/-.

## MEDIUM RANGE ELECTROLYTICS

 Axial leads: $50 / 501 / 9 ; 100 / 251 / 9 ; 100 / 502 / 6 ; 250 / 252 / 6$; 4/-; 1000/50 6/-; 2000/25 6/-
## SMALL ELECTROLYTICS

Axial leads: $47 / 10: 4.7 / 25$; $5 / 50$ 1/- ea. $10 / 10 ; 10 / 25$; $10 / 50,33 / 10 ; 50 / 101 /-$ ea. $22 / 25 ; 25 / 50 ; 47 / 25 ; 100 / 10 ;$
$220 / 101 / 3$ ea. 220/10 1/3 ea.
ENAMELLED COPPER WIRE, even No S. W.G only. 2oz reels $16-22$ ©.W.G. 4/3; 24-30 B.W.G. $5 /-: 32-34$ S.W.G 5/6: 36-38 S.W.G. 6/3. 4 oz reels 18-22 B.W.G. only 7/6.

## PEAK SOUND PRODUCTS ENGLEFIELD KITS

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Stereo amplifier in modular kit form 12 watts per channe \&88/0/-; Cabinet kit only $\mathbf{4 6}$. These prices nett. As recently reviewed in Hi Fi Bound.
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Designed by Peter Baxandall.
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speaiker Unit Kit $\mathbf{2 1 8 / 1 2 / - n e t t ; ~}$
built $£ 19 / 8 / 6$ nett.

## MAINLINE AMPLIFIER KITS

RCA/Gos designed main amplifier kits
Input sensitivity $500-700 \mathrm{mV}$ for full output into $8 \Omega$.
Power Kit price Suitable unres.
$\left.\begin{array}{ccc}\text { Power } & \begin{array}{c}\text { Kit price } \\ \text { including } \\ \text { components }\end{array} & \begin{array}{c}\text { Suitable unreg. } \\ \text { power supply }\end{array} \\ & 168 /- \text { kit }\end{array}\right]$

## 30 WATT BAILEY AMP. PARTS

 Sensitivity 1.2 V for full output into $8 \Omega$Transistors and PCB for one channel $\mathbf{\text { 57/5/6 }}$
Transigtors and PCB for two channols si4/11/-
Capacitors and resistors (metal oxide) 40/-per channel

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$5 \%$ full range E 24 values; $400 \mathrm{~mW}: 2.7 \mathrm{~V}$ to $30 \mathrm{~V} 3 / 9$ each Clip to increase 1.5 w rating to 3 wats to $75 \mathrm{~V} 12 j-\mathrm{each}$

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POTENTIOMETERS, long spindles
Double wiper ensuree minimum noise level
Single gang linear $220 \Omega$, to $2 \cdot 2 \mathrm{M} \Omega$
Dual gang linear $4.7 \mathrm{~K} \Omega$ to $2 \cdot 2 \mathrm{M} \Omega$
$\begin{array}{ll}\text { Dual gang linear } & 4.7 \mathrm{~K} \Omega \text { to } 2.2 \mathrm{M} \Omega \\ \text { Dual gang log } & 4.7 \mathrm{~K} \Omega \text { to } 2.2 \mathrm{M} \Omega\end{array}$
$\begin{array}{ll}\text { Log/antilog } & 10 \mathrm{~K}, 47 \mathrm{~K}, 1 \mathrm{M} \Omega \text { only }\end{array}$
Any type with $\frac{1}{2}$ A D.P. mains switch, extra
Please note: only decades of 10,22 and 47 are available within ranges quoted.

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| $\begin{aligned} & \mathrm{OA} 2 \\ & \mathrm{OA} 3 \end{aligned}$ | 0.38 | 6 AR5 | 0.35 | 6E57 | 0.35 | 7 Y 40.6 | First Quallty |  |  | Fully Guaranteed |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 0.45 | 6AR6 | 0.40 | 6EW6 | 0.65 | 98W6 0.50 |  |  |  |  |  |  |  |
| $\mathrm{OB}^{\text {O }}$ | 0.35 | 6AS5 | 0.35 | 6F1 | 0.70 | 10 C 20.50 |  |  |  |  |  |  |  |
| OB3 | 0.60 | 6AS7G | 0.80 | 6 F 5 | 0.50 | 10D1 0.50 |  |  |  |  |  |  |  |
| OC3 | 0.38 | ${ }^{64 T 6}$ | 0.30 | ${ }_{6}^{6 \mathrm{FbG}}$ | 0.30 | 10D2 0.40 |  |  |  |  |  |  |  |
| OD3 | 0.35 | 6av6 | 0.25 | ${ }_{6}^{6 \mathrm{Fll}}$ | 0.38 | $10 \mathrm{Fl}{ }^{10.90}$ |  |  |  |  |  |  |  |
| Ib3GT | 0.38 | 6AV6 | 0.30 | 6 F 13 | 0.38 | ${ }_{10 \mathrm{FF}}$ |  |  |  |  |  |  |  |
| $\begin{aligned} & \text { ML4 } \\ & \text { IR4 } \end{aligned}$ | 0.20 0.85 | ${ }_{6}^{6 A W} 88$ | 0.55 | ${ }_{6}^{6 F 14}$ | 0.85 | $\begin{array}{lll}10 \mathrm{F1} & 0.45 \\ 10 \mathrm{~L} 1 & 0.45 \\ & 0.45\end{array}$ |  |  |  |  | BRA |  |  |
| IR5 | 0.85 | 6BE6 | 0.30 | ${ }^{6 F 18}$ | 0.45 | $10 \mathrm{LD11} 0.80$ |  |  |  |  |  |  |  |
| IS4 | 0.27 | ${ }_{68 \mathrm{BF5}}^{68}$ | 0.80 0.50 | ${ }_{6}^{6 \mathrm{~F} 23}$ | ${ }_{0}^{0.80}$ | ${ }_{10 \mathrm{P} 13} 10.55$ |  |  |  |  |  |  |  |
| IT4 | 0.25 | ${ }^{6} \mathrm{BH} 6$ | 0.45 | 6 F 25 | 0.75 | ${ }_{12 \mathrm{AB5}}{ }^{19.60}$ | EL |  |  |  |  |  |  |
| IV4 | 0.27 | 6BJ6 | 0.45 | 6F26 | 0.35 | $12 \mathrm{AC6} 0.40$ |  |  |  |  |  |  |  |
| IU5 | 0.50 | 6BK7A | 0.55 | 6 F 28 | 0.60 | 12AL5 0.45 |  |  |  |  |  |  |  |
| IV2 | 0.45 | 6 BL 7 GT | A | ${ }^{654} 4$ | 0.50 | 12AQ5 0.43 | 30AE3 0.40 | 50 CD | 91.65 | DK92 | 0.50 | ECC84 |  |
| ${ }_{2 \mathrm{LX} 2 \mathrm{~B}}$ | 0.40 0.35 |  | 0.65 0.43 | ${ }_{6 J 7}^{655 C T}$ | 0.30 | 12AT66 0.30 | ${ }_{30 \mathrm{Cl}}^{30 \mathrm{Cl}} 0.0 .30$ | 50L6C | 0.50 | DK96 | 0.42 | ECC85 | . 60 |
| ${ }_{3 \text { 2 }}$ | 0.35 0.35 | 6BN5 | 0.43 0.40 | ${ }_{6}^{656} 6{ }_{\text {GT }}$ | 0.45 | $\begin{array}{ll}\text { 12AT7 } & 0.33 \\ 12 A U 6 \\ 0.30\end{array}$ | $\begin{array}{ll}30 \mathrm{C} 15 & 0.80 \\ 30 \mathrm{C} 17 & 0.85\end{array}$ | ${ }_{90 A V}^{85 A}$ | 0.40 2.50 | DL96 | 0.42 0.65 | ECC888 | 0.40 |
| 3328 | 2.15 | 6BQ5 | 0.25 | .6K7 | 0.35 | $12 \mathrm{~A} \mathrm{Cl}^{0} 0.30$ | $\begin{array}{ll}30 \mathrm{Cl} 18 & 0.75\end{array}$ | 90 Cl | 0.60 | DY86 | 0.33 | ECCOI | 0.20 |
| $3 \mathrm{BP1}$ | 2.75 | 6BR8 | 0.85 | 6K8G | 0.35 | 12AV6 0.33 | ${ }^{30 \mathrm{~F} 5} 00.85$ | 90 CV | 1.25 | ${ }^{\text {DY87 }}$ | 0.35 | ECC189 0 | 0.60 |
| $3 \mathrm{Q4}$ | 0.40 | 6BS7 | 1.35 | ${ }^{6 \mathrm{~K} 23}$ | 0.55 | 12AV7 0.50 | 30FLl 0.70 | 807 | 0.50 | DY802 | 0.50 | ECF80 | 0.35 |
| 3 S 4 | 0.35 | ${ }^{68 W 6} 6$ | 0.85 | ${ }^{6 \mathrm{~K} 25}$ | 0.75 | $12 \mathrm{AX7} 70.30$ | 30FL12 0.93 | 812A | 3.50 | E55L | 2.75 | ECF82 |  |
| 3 V 4 | 0.45 | ${ }^{6 B W 7}$ | 0.70 | ${ }^{6 L 6 G T}$ | 0.45 | 12AY7 0.70 | $30 \mathrm{FL14} 0.75$ | 813 | 3.75 | E88CC | 0.85 | ECF86 |  |
| 5 S 4 GY | 0.60 | 6 BX 6 | 0.25 | 6L7 | 0.40 | 12 B 44 A 0.55 | $30 \mathrm{L1} \quad 0.40$ | 866A | 0.75 | E130L | 5.00 | ECF804 |  |
| ${ }^{5044 G}$ | 0.33 | ${ }_{68 \mathrm{Br}}$ | 0.35 | ${ }^{6 L 18}$ | 0.45 | 12BA6 0.35 | $30 \mathrm{L15} 50.85$ | 5642 | 0.65 | E180F | 0.95 | ECH42 |  |
| ${ }_{5}^{50} 4 \mathrm{GGB}$ | 0.42 | 6C4 6C5GT | 0.33 0.40 | ${ }_{6 N \mathrm{~L} 7 \mathrm{GT}}$ | 0.40 | $\begin{array}{ll}\text { 12BA7 } \\ \\ 12 \mathrm{BEF} & 0.35 \\ 0.35\end{array}$ | $\begin{array}{ll}30 \mathrm{L17} & \mathbf{0 . 8 0} \\ 30 \mathrm{P} 12 & 080\end{array}$ | 6080 | 1.50 | eabcio |  | ECH81 | 0.30 |
| 5Y3GT | 0.32 | 6 CA 4 | 0.28 | $6 \mathrm{P1}$ | 0.60 | ${ }^{12 \mathrm{BH} 7} 70.40$ | $\begin{array}{ll}30 \mathrm{Pr12} & 0.80 \\ 30 \mathrm{P} 19 & 0.80\end{array}$ | ${ }_{61468}^{6146}$ | 2.50 | EAF42 | 0.55 | ECH84 | 0.45 |
| 5 23 | 0.50 | 60a7 | 0.50 | 6 P 28 | 0.65 | 12BY7 0.55 | 30PL1 0.70 | 6360 | 1.25 | EBC41 | 0.55 | ECL80 | 0.40 |
| 5Z4G | 0.40 | ${ }^{6 C B 6}$ | 0.30 | 6 6 7 | 0.40 | 12 K 50.55 | 30PL13 0.93 | 6939 | 2.15 | EbC81 | 0.30 | ECL81 | 0.45 |
| 6/30L2 | 0.75 | 6CD6GA |  | 6SA7 | 0.40 | 12K7G10.35 | 30 PL 140.90 | 7199 | 0.75 | EBF80 | 0.40 | ECL82 | 0.35 |
| 6AB4 | 0.35 |  | 1.15 | ${ }_{6 S 67}^{6597}$ | 0.35 | 1207 CH 0.30 | ${ }^{3543} 0.55$ | 7360 | 1.80 | EBF83 | 0.40 | ECL83 | 0.65 |
| 6AF4A | 0.50 | ${ }_{6}^{6 C G 7}$ | 0.50 | 6SK7 | 0.35 | ${ }^{12 S R} 70.35$ | ${ }^{35155} 50.75$ | 7586 | 1.25 |  | 0.30 | ECL84 | 0.55 |
| 6AG7 6 6 | 0.40 0.50 | ${ }_{6}^{6 \mathrm{CLH} 6}$ | 0.0 .55 | $\begin{aligned} & \text { 6SL7GT } \\ & \text { 6SN7GT } \end{aligned}$ | 0.35 | $\begin{array}{ll}\text { 1487 } & 0.80 \\ \text { 20D1 }\end{array}$ | $\begin{array}{ll}35 \mathrm{BV} & 0.65 \\ 3 \overline{\mathrm{C}} 5 & 0.40\end{array}$ | 7895 9002 | 1.25 | ${ }_{\text {EC83 }}^{\text {EC53 }}$ | 0.50 | ECLS5 | 0.55 0.40 |
| 6AJ8 | 0.30 | 6CW4 | 0.63 | 6SQ7 | 0.40 | ${ }_{20 \mathrm{LL}}^{20} 10$ | $\begin{array}{ll}35 \mathrm{D5} & 0.70\end{array}$ | 9003 | 0.50 | EC88 | 0.60 | ECLL 800 |  |
| 6AK5 | 0.30 | ${ }_{6}^{6 C Y}{ }^{6} 7$ | 0.65 | ${ }^{63 R 87}$ | 0.40 | ${ }^{2091} 0$ | $35 \mathrm{L6GT} 0.50$ | AZ1 | 0.48 | EC90 | 0.38 | - 1 | 1.50 |
| 6AK6 | 0.57 | 6D3 | 0.45 | ${ }^{6 T 8} 8$ | 0.35 | $20 \mathrm{P4} 4.10$ | $35 \mathrm{W4} 0.30$ | AZ31 | 0.55 | EC92 | 0.85 | EF37A | 0.60 |
| 6 6AL3 | 0.43 | 6DC6 | 0.75 | ${ }^{\text {6UU } 4 \text { GT }}$ | 0.60 | ${ }^{20 \mathrm{~Pb}} \quad 1.20$ | $\begin{array}{ll}3573 & 0.60\end{array}$ | ${ }^{\text {CBL1 }}$ | 0.80 | ${ }_{\text {EC93 }}$ | 0.50 |  | 0.40 |
| $\begin{aligned} & \text { 6ALS } \\ & \text { 6AMS } \end{aligned}$ | $\begin{aligned} & 0.20 \\ & 0.32 \end{aligned}$ | ${ }^{6 \mathrm{DKG}} \mathrm{E}$ | $\begin{aligned} & 0.48 \\ & 0.83 \end{aligned}$ | $\begin{aligned} & 6 \mathbf{U} 8 \mathrm{~A} \\ & 6 \mathrm{X} 4 \end{aligned}$ | $\begin{aligned} & 0.40 \\ & 0.80 \end{aligned}$ | $\begin{aligned} & 2505 \\ & 25 \mathrm{~L} 6 \mathrm{GT} \\ & 0.50 \\ & 0.45 \end{aligned}$ | 3524 G <br> 35 O | ${ }_{\text {CYL31 }}$ | 0.90 0.35 | ECC40 | 2.25 0.60 | EF40 | 0.50 0.65 |
| 6amo | 0.33 | 6DS4 | 0.76 | 6x5GT | 0.35 | $25 \mathrm{Z4G} 0.30$ | $\begin{array}{lll}50 \mathrm{~A} 5 & 0.70\end{array}$ | DAF96 | 0.42 | ECC81 | 0.83 | EF42 | 0.70 |
| 6AQ5 | 0.35 | 6EA8 | 0.58 | 6x8 | 0.55 | 25Z6GT 0.65 | 5085 0.45 | DF96 | 0.42 | Ecc82 | 0.30 | EF80 | 0.25 |
| 6AQ6 | 0.55 | 6EE7 | 0.30 | $6 \mathrm{Y6G}$ | 0.65 | 30A5 0.45 | 50C5 0.40 | DK40 | 0.55 | ECC83 | 0.30 | EF83 | 0.55 |

## TRANSISTORS

|  |
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|  | 0.85 | EZ40 0.45 | PCL88 0.90 | U78 |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | 0.30 | EZ41 0.45 | PCL800 0.93 | U191 | 5 |
| EF89 | 0.28 | EZ80 0.25 | PCL801 0.70 | U201 | 0.35 |
| EF91 | 0.33 | EZ81 0.28 | PD500 1.50 | U28 | 0.40 |
| F92 | 0.40 | GY501 0.80 | PF86 0.60 | U28 | 0.40 |
| F95 | 0.30 | Gz30 0.40 | PF818 0.85 | U301 | 0.40 |
| F97 | 0.65 | ZZ31 0.83 | PFL2000.70 | U403 | 0.50 |
| P98 | 0.65 | Z332 0.48 | PL33 0.85 | U404 | 40 |
| F183 | 0.30 | 0.70 | PL36 0.55 | U801 | 0 |
| EF184 | 0.85 | Z34 0.80 | PL81 0.50 | UAB |  |
| EF800 | 1.00 | HABC80 | PL82 0.45 |  | 0.35 |
| EF804 | 1.25 | 0.45 | PL83 0.45 | UAP4 | 0.50 |
| EK90 | 0.30 | HK90 0.35 | PL84 0.40 | UAF4 | 0.55 |
| EL33 | 1.25 | KT66 1.70 | PL302 0.80 | UBC41 | 0.50 |
| EL34 | 0.50 | KT88 1.75 | PL504 0.80 | UBC81 | 0.40 |
| EL36 | 0.50 | N78 1.15 | PL508 0.90 | UBF80 | 0.40 |
| EL41 | 0.55 | PABC800.40 | PL509 1.30 | UBF89 | 0.35 |
| EL42 | 0.58 | PC86 0.60 | PL509 1.30 | UBL1 | 0.50 |
| EL81 | 0.55 | PC88 0.60 | PL801 0.80 | UBL2 | 0. |
| L83 | 0.42 | PC97 0.50 | PM84 0.50 | UC92 | 0.35 |
| L84 | 0.25 | PC900 0.48 | $\begin{array}{ll}\text { PY31 } & 0.30\end{array}$ | JCC8 | 0.40 |
| EL85 | 0.48 | PCC84 0.40 | PY33 0.63 | UCF8 | 0.55 |
| EL86 | 0.40 | PCC85 0.40 | PY80 0.35 | UCH2 | 0.60 |
| EL90 | 0.35 | PCC88 0.55 | PY81 0.30 | UCH4 | 0.70 |
| EL95 | 0.35 | PCC89 0.50 | PY82 0.30 | UCH43 | 0.75 |
| EL360 | 1.15 | PCC189 0.55 | PY83 0.88 | UCH81 | 0.35 |
| EL803 | 1.00 | PCC805 0.85 | PY88 0.40 | UCL81 | 0.80 |
| EL821 | 0.55 | PCC806 9.80 | PY500 1.00 | UCL82 | 0.35 |
| EL822 | 0.90 | PCF80 0.80 | PY800 0.50 | UCL83 | 0.60 |
| ELL80 | 0.75 | PCF82 0.85 | PY801 0.50 | UF41 | 0.60 |
| EM34 | 0.90 | PCF84 0.50 | PZ30 0.35 | UF42 | 0.60 |
| EM71 | 0.75 | PCF86 0.60 | QQV2-6 2.15 | UF43 | 0.60 |
| M80 | 0.40 | PCF87 0.85 | 3-10 | UF80 | 0.85 |
| EM81 | 0.60 | PCF8010.50 | 1.25 | UF85 | 0.40 |
| EM84 | 0.35 | PCF802 0.50 | -20A | UF89 | 0.35 |
| EM85 | 1.00 | PCF805 0.75 | 5.25 | UL41 | 0.65 |
| EM87 | 0.55 | PCF806 0.70 | TT21 2.65 | UL84 | 0.30 |
| EN91 | 0.35 | PCF808 0.75 | TT22 2.80 | UM84 | 0.20 |
| EY51 | 0.40 | PCH2000.70 | U18/20 0.75 | UY1N | 0.50 |
| EY80 | 0.45 | PCL81 0.50 | U25 0.75 | UY11 | 0.65 |
| EY81 | 0.40 | PCL82 0.85 | U26 0.75 | UY41 | 0.45 |
| EY83 | 0.55 | PCL88 0.65 | U31 0.45 | UY82 | 0.50 |
| EY86 | 0.40 | PCL84 0.45 | U37 1.50 | UY85 | 0.30 |
| EY | 0.48 | PCL85 0.40 | U52 0.33 | W729 | 0.60 |
| EY88 | 0.43 | PCL86 0.45 | U76 0.30 | Z803U | 0.90 |

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| BY127 | 600 p.i.v. | 1 Amp | 15 pp |
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