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Wireless World

Radio · **Electronics** · **Television**

'FORTY-FIFTH YEAR OF PUBLICATION

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Wireless World

NOVEMBER 1955

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RADIO, ELECTRONICS, TELEVISION

Managing Editor: HUGH S. POCOCK, M.I.E.E. Editor : H. F. SMITH

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WIRELESS WORLD

NOVEMBER, 1955



ALVES, TUBES & CIRC

35. HIGH QUALITY SOUND REPRODUCTION WIDE RANGE OF CIRCUITS

The high quality reproduction obtainable with modern audio valves at low cost has already been demonstrated by the circuit for a 5-valve 10-watt amplifier introduced last year. The '5-10' is now a firm favourite because of the easc with which it can be constructed and because it can be driven directly by many of the more popular and inex-

pensive types of crystal pickup. Many constructors will, however, be interested in building a preamplifier for the '5-10' in order to use a pickup giving 40mV output, perhaps a magnetic type or a crystal pickup loaded for output proportional to stylus velocity. Suitable single-valve units of straightforward construction have been designed by Mullard engineers. There are two versions of the basic design. The first is referred to as preamplifier 'A', and is the one required by those who have already made a '5-10' in which the treble, bass, and volume controls are included in the main amplifier. The second version is known as preamplifier 'B', and is more suitable for those who are only just going to build the amplifier. Here the tone and volume controls are included in the preamplifier itself, giving more convenient access to the controls-particularly important if the equipment is to be assembled into a cabinet. The tone and volume controls in preamplifier 'B' are similar to those which can be incorporated in the main amplifier (as in the original design) but with the appropriate changes to component values.

Both 'A' and 'B' are suitable for microphone input, or tape input with 30dB bass pre-emphasis at 120c/s.

Point-to-point wiring diagrams and chassis drawings are given in the booklet 'High Quality Sound Reproduction' for preamplifiers 'A' and 'B' and for the two corresponding versions of the '5-10' itself. The booklet also gives details of minor improvements and modifications made to the original amplifier circuit. Chiefly they are intended to take advantage of the high quality of the output transformers which have become available commercially and to give increased stability at the high and low frequency ends of the response. The rated output of the amplifier remains at 10 watts, but the overload point (onset of clipping with



NEW PUBLICATIONS

'High Quality Sound Reproduction' is available from radio retailers price 3s. 6d. It contains 48 pages, of which 44 are devoted to the designs described elsewhere on this page. There are 24 pages of information not previously published, including 14 pages of constructional drawings. The pages are 10 ins. by 8 ins.

The 'Mullard Maintenance Manual' is now on sale at radio retailers price 10s. 6d. It provides the information of most use to the service engineer for repairing and maintaining radio and television sets and audio amplifiers, and includes data on valves and tubes for maintenance purposes as well as on the latest types. Constructors and designers will also find this manual extremely useful as a quick reference guide to data. For the complete information required in designing new equipment the Mullard Technical Handbook (available on a subscription basis) should be consulted.

To help home constructors starting to experiment with transistor circuits a new leaflet is available free of charge entitled 'Junction Transistors for the Home Constructor'. It contains eight circuits using Mullard OC70 and OC71 junction transistors, which are now readily available. Background information is included for those who would like to go more deeply into the subject. normal or low loading is employed in the output stage. The EZ81 has the same pinning as the EZ80 originally specified, but the *total* limiting resistance at each anode of the EZ81 must be at least 310Ω in this circuit.

Other Mullard circuits designed for high quality sound reproduction have been published recently in Wireless World. A 20-watt amplifier, sufficient for a small to medium-sized hall, was described in the May and June issues. The use of distributed load operating conditions in the EL34 push-pull output stage gives a total harmonic distortion of less than 0.05% at the 20 watts nominal output. A 3-valve preamplifier suitable for this and similar amplifiers was described in the July Wireless World, and the August issue contained an article on an f.m. tuner unit which can provide the input for the 10-watt or 20-watt amplifier. Point-to-point wiring diagrams and chassis drawings for these circuits, in addition to reprints of the Wireless World articles, are included in 'High Quality Sound **Reproduction'.**

À 3-valve 3-watt amplifier has been designed for those who re-

quirereasonably high quality at low cost and with the simplest possible construction. The total harmonic distortion is 1.5% at 3 watts; the sensitivity is 100mV, and the amplifier is suitable for all types of crystal pickup. This amplifier can be driven by the f.m. tuner unit via a 5:1 attenuator. An EZ81 must be included to provide power for the f.m. tuner, and the mains transformer must then be rated at 90mA or 100mA. Preamplifier 'A', primarily designed for the 10-watt amplifier, is suitable here also; a 5:1 attenuator is required to preserve the original basic sensitivity. Without the attenuator the sensitivities are increased by a factor of 5. Details of the 3-valve 3-watt amplifier are available only in the form of a leaflet (free, from the address below).



The free leaflets and further information can be obtained from

MULLARD LTD., Technical Service Dept., Century House, Shaftesbury Avenue, London, W.C.2.

MVM 338



NOVEMBER 1955

Television Complications

BRITISH television, in its original form, grew up pretty much according to plan and, in technical matters at least, the situation was always fairly well in hand. But subsequent developments, and particularly the incursion into another frequency band, have shown that it is almost beyond the powers of human foresight to envisage all the possibilities of trouble. That is why we are glad to see that the B.B.C. experimental transmissions in colour, which began on October 10th, were embarked upon with due caution and with no suggestion that the system used will be chosen for a regular service. According to the Corporation's statement, the so-called British N.T.S.C. system "*might* prove suitable" (our italics).

The British Radio Equipment Manufacturers' Association, in welcoming the start of the colour experiments, showed equal caution, and expressed the view that at least two years would elapse before any decision could be made.

Naturally enough, the industry is anxious that the general public should not be under the mistaken impression that colour television is likely to start in the immediate future. There is every reason, however, why the problems of colour should, in the meanwhile, be freely discussed in technical circles. Indeed, that applies to all aspects of television, and we sometimes think the official mind is unduly shy of the limelight.

An example of what seems to be unnecessary reticence has arisen over the transmission characteristics of the new London station at the Crystal Palace. A B.B.C. statement has just been issued saying that it is to work on the vestigial sideband method on vision, with the upper sideband partially suppressed; the present Alexandra Palace station works on double sideband. This change was generally expected and, indeed, was foreshadowed in *Wireless World* a couple of years ago; why the delay in making the announcement?

Another of the unexpected difficulties of television has come to light in connection with t.r.f. receivers, some of which will be rendered obsolete by the change mentioned in the last paragraph. As described in an article on p. 526, the use of these receivers, in conjunction with a Band III convertor, has been found to produce serious interference with nearby receivers tuned to Band I. Though we have published warnings against various difficulties in the use of convertors, we must admit to having overlooked this particular trouble.

Gallons Into Quart Pots

WE are constantly being reminded that the world's supplies of such vital commodities as coal and oil are running low. But complete exhaustion of our resources is never quite in sight, and, if and when the worst happens, nuclear energy is confidently expected partly at least to make good the deficiency.

Not so with the radio communication spectrum, in which there are no rich seams yet undiscovered and no new "gushers" waiting to be tapped. The full extent of our national resources is known precisely; these resources are inelastic and all that can be done is to use the communication channels we have to the best effect.

One method, which is of great practical significance at present, is to reduce the channel width. Many allocations were made on what now seems to be an unnecessarily lavish basis. A case in point is that of the mobile radio-telephone service, for which the channel width as at present allocated is 100 kc/s. The possibility of reducing this to 50 kc/s was considered in a report made to the Postmaster-General some months ago, but no decision was reached. Now, as reported on p. 527, a demonstration has been given of interference-free working in adjacent channels only 25 kc/s wide—an achievement twice as great as the proverbial one of getting a quart into a pint pot.

This excellent example of channel conservation has been brought about by the extension of straightforward and well-tried methods, and so there seems to be no technical reason why 25-kc/s channels should not be generally adopted in the mobile service, for which there is an acute shortage of space. The difficulties appear to be mainly economic, though it is understood the new 25-kc/s equipment is not significantly more costly.

VOL. 61 No. 11

Interference from Band-III Convertors

By F. HOWARD STEELE

Radiation on 45 Mc/s

HE advent of television broadcasting in Band III appears to add yet another source of interference to the B.B.C. transmissions in Band I. The trouble arises from Band-I t.r.f. set owners who have converted their sets to receive Band-III transmissions by adding one of the popular "universal" convertors to the front end of their receivers. These convertors usually consist of an r.f. stage followed by an oscillator/mixer, the oscillator being tuned so that the convertor output frequency is the same as the local B.B.C. transmission in Band I.

This I.T.A.-modulated Band-I signal is then amplified through the t.r.f. stages of the receiver until in the region of the detector a signal of several volts in amplitude is obtained. Unless the screening of the receiver is unusually efficient, a few millivolts of this I.T.A.-modulated Band-I signal is picked up by either the Band-I or Band-III aerial leads and is re-radiated. Because of local oscillator drift in the Band-III convertor this radiated signal is not of constant frequency, and variable frequency beats are produced with the B.B.C.'s Band-I transmissions.

The effect on the screen of television receivers operating on Band I anywhere within 100 yards or so of the offending "viewer" is extremely irritating. A vicious herring-bone pattern appears on the screen which is constantly on the move, starting first one way and then the other as the re-radiated signal drifts above and below the frequency of the local B.B.C. transmission in Band I.

Additional Complications

The cause of this interference would seem difficult enough to cure as it is, without two more contributory factors which make it well nigh impossible. The first of these additional complications is the number of t.r.f. receivers which are of the a.c./d.c. type and therefore cannot be properly earthed. The second is the current practice of using unbalanced coaxial cable as the downlead from an aerial which is inherently a balanced device. This prevents one from considering the downlead as a simple transmission line.

The last-mentioned complication can be remedied, of course, by fitting a balance-to-unbalance transformer at the aerial end of the downlead, but if this is done without careful thought impedance matching problems arise. Theoretically, if correct matching is to be obtained, one should use a quarter-wave line transformer between the end of the feeder and the upper element of the dipole and a three-quarter wavelength line between the end of the feeder and the lower element of the dipole. Apart from the complication of so doing, however, it would tend to make the collection of scrap iron we have on our chimneys these days look even more untidy than it does at present.

Should any readers consider doing this, however, it

should be remembered that the mechanical length of a piece of coaxial 75-ohm feeder is shorter than its electrical length, because of the slower propagation rate in this type of cable. The transmission factor of in solid Polythene insulated coaxial is approximately 0.66 and the matching stubs should be shorter than the equivalent free-space wavelength by this factor.

The problem of providing a satisfactory r.f. earth at the receiver input terminals is more difficult. In theory the length of the earth wire should be a multiple of $\lambda/2$ where λ = wavelength of local B.B.C. transmission in Band I. In practice, it would appear impossible to calculate this length, due to the indeterminate stray capacitances to the surroundings of this earth wire and the difficulty of knowing exactly where "earth" is at the far end. A semi-practical solution is to insert a series LC circuit in the earth lead and by varying the L or the C attempt to tune the earth wire to an exact half wavelength. If this can be done (and so far the author has failed to do so) then the receiver input will be at r.f. earth potential so far as the Band-I transmission is concerned and, in conjunction with a balance-to-unbalance modification to the aerial system, one will be well on the way to solving the re-radiation problem.

Having stopped the t.r.f. receiver "bouncing up and down at the end of its earth wire" at I.T.A.modulated Band-I frequency, the next problem is how to stop the spurious Band-I radiation from the t.r.f. receiver being passed back up the aerial feeders and being re-radiated by either the Band-I or Band-III aerials.

Considering the Band-III aerial first, the solution in theory is not too difficult. All that is required is a high-pass filter to be fitted in the aerial feeder as far away from the t.r.f. receiver as possible. This filter should look like 75 ohms when viewed from either the aerial or the receiver if a mis-match is not to occur and it should ideally have zero insertion loss at Band-III frequencies and an infinite insertion loss at Band-I frequencies. In practice a simple π -type filter with a capacitor as the series element appears to work reasonably well, but the calculated values of inductance in the shunt arms are so small as to make their construction largely a matter of trial and error. (To obtain the required 0.148μ H one or two turns, $\frac{1}{4}$ in diameter air-spaced, are necessary.) In addition, the two inductors need to be separated by an earthed screen (through which the series capacitor, about 12pF, passes) to avoid mutual coupling between them. The whole filter is then placed inside a fully screened box. Such a filter has a designed cut-off frequency of approximately 90 Mc/s and a characteristic impedance of 75 ohms-we hope!

In practice such a filter makes a substantial reduction in the level of radiated interference but is not in itself a complete cure by any means. One is

still not on speaking terms with one's immediate neighbours

Stopping the Band-I aerial re-radiating is more difficult, and so far no satisfactory cure has been found. One cannot put a high-pass filter or even a band-stop filter in this downlead because, being a two-way device, it precludes the possibility of observing the Old Lady of Shepherd's Bush when one wants to.

Some Band-III convertors on the market shortcircuit the Band-I aerial when it is not in use, but this discontinuity, of course, tends to set up standing waves in the Band-I aerial feeder with consequent radiation. A better solution would appear to be terminating the unused aerial feeder in 75 ohms and screening the termination.

It will be gathered from the above that the author owns a t.r.f. receiver and a Band-III convertor of the type referred to. Both the receiver and the convertor are fully screened and yet he must confess, however unknowingly at the time, to causing Band-I interference on the opening night of the I.T.A. transmissions—a fact which a kind but suspicious neighbour was not slow to suggest the following day. Since that time considerable thought has gone into ways and means of suppressing this interference, which in the author's location appears to be a very real problem. A considerable reduction in the level of interference has resulted, but the interference pattern is still visible on receivers up to 100 yards away.

The only foolproof solution would appear to be the

conversion of the t.r.f. receiver to a superhet, using a commercial turret tuner as a front end and dropping the existing 45-Mc/s r.f. strip down to the new preferred British i.f. of 34.65 Mc/s. As the oscillator in such tuners is usually above the Band-I frequency (to avoid harmonic breakthrough) this inverts the relative position of the sound and vision channels and the sound i.f. comes out at 38.15 Mc/s. This, fortunately, is within the range of adjustment of the majority of existing sound receiver circuits and no modification is necessary.

Pending this rather drastic step, the author has not dared to tune in to the I.T.A., and is all the more infuriated to find that others are now causing exactly the same interference as he did on the opening night. It is small consolation to be able to show suspicious neighbours who "just drop in for a chat" that he, too. is suffering the same interference as they are.

Unless anyone can suggest a lasting, immediate and hundred-per-cent effective cure, it is up to all owners of t.r.f. receivers plus convertors to check without delay whether they are radiating this interference. Moreover, radio dealers should be discouraged from selling these Band-III convertors to any t.r.f. set owner unless he has checked that no interference will result. To have bought a convertor, to have sniffed the rabbit (however young and immature) and not to be able to watch his progress is almost intolerable, and yet it seems the only way if one is to remain at peace with those who live nearby.

Narrower Mobile Radio Channels

BAND III is now in the process of being cleared for television purposes and when the job is completed the space left for private mobile radio users will be only about 7 Mc/s. This imposes a severe restriction on the development of mobile radio which can only be overcome by reducing channel widths. An official body has already recommended that the present 100-kc/s channel width should be reduced to 50 kc/s* but it now seems that the belt can be tightened even more. Pye Telecommunications have recently demonstrated a new two-way mobile radiotelephone, known as the "Ranger," designed for channels as narrow as 25 kc/s. This, of course, makes possible four channels in the space occupied by one at present.

With such close spacing the problem of cross-talk between channels becomes really acute, and it is essential that good frequency stability shall be maintained in the equipment. The Pye transmitters and receivers are crystal-controlled, of course, and the crystals themselves are cut in a manner which gives high stability operation (AT-cut crystals). The transmitter consists of an oscillator followed by three frequency multipliers and a power amplifier, with a modulator comprising a phase splitter and push-pull output stage. An r.f. output power of 4-6 watts is obtained and the frequency drift is claimed to be less than $\pm 0.002\%$. The receiver is a double superheterodyne with two crystals and eleven valves and is provided with noiselimiting and a.g.c. circuits. Sensitivity is $1\mu V$ for 100mW output, while the frequency drift here is claimed to be less than ± 0.0025 %.

The equipment will work anywhere in the frequency range 25-174 Mc/s on either single- or double-

* "Report of the Mobile Radio Committee," H.M.S.O., price 9d; summarized in "Mobile Radio," Wireless World, May, 1955. frequency simplex, and is operated by a "press-totalk" switch on the hand microphone. Transmitter, receiver and power supply are housed in a steel case intended for fitting under the dashboard of a motor vehicle, as shown in the photograph below.

In the demonstration of the equipment, four mobile units in motor vehicles were receiving from four 15watt fixed transmitters in adjacent 25-kc/s channels (actually 161.725, 161.75, 161.775 and 161.8 Mc/s), the mobile transmitters and corresponding fixed receivers being spaced from these by 4.5 Mc/s (at 157.225, 157.25, 157.275 and 157.3 Mc/s respectively). In spite of this close spacing there appeared to be no crosstalk between channels except a slight trace when the mobile units had been driven very close to the fixed transmitter aerials.



WORLD OF WIRELESS

British N.T.S.C. Tests + H.P. and Retail Sales + Engineering Opportunities

Colour Television Tests

THE promised television test transmissions in colour began from Alexandra Palace on October 10th. The system being used in these initial tests, which are being radiated outside normal programme hours, is the British adaptation of the American N.T.S.C. system (see our August issue). Other systems may be tried later on.

These tests, which are being carried out in agreement with the Television Advisory Committee, are mainly to ascertain whether a truly compatible system should be the final objective. Viewers in the London area will have the opportunity of finding out if there is any appreciable degradation of the monochrome picture.

New Television Stations

TWO new television stations, both operating in Channel 4 (61.75 Mc/s vision and 58.25 Mc/s sound) have been brought into service by the B.B.C. during the past month.

The first, a 1-kW temporary Channel Islands transmitter at Les Platons, Jersey, at present has to rely on the reception of transmissions from the temporary station at North Hessary Tor. The permanent Devon station will be in use early next year. The signals are picked up at a receiving station at Torteval, Guernsey, and conveyed to Jersey by a radio link.

The other new station is the permanent transmitter at Meldrum, 20 miles north-west of Aberdeen. It has an e.r.p. of 12 kW and replaces the temporary transmitter at Redmoss, which has been in use for the past 12 months.

Both stations use horizontal polarization.



OPTICAL ASSEMBLY of the three-tube colour camera which, with its associated equipment, has been supplied by Marconi's for the B.B.C. colour television tests. The film and slidescanning equipment being used in the tests was developed and built by the B.B.C. Research Department. Rear-Admiral K. H. T. Peard speaking at the farewell dinner to Rear-Admiral Sir Philip Clarkewhom he succeeds as director of the naval electrical department. Sir Philip has been nominated for a second year as president of the Brit. I.R.E.



Retail Sales

THE increase in the hire-purchase deposit on broadcast receivers introduced at the end of July is reflected in the survey of retail sales during August, prepared by the British Radio Equipment Manufacturers' Association. It will be seen from the table, in which the hire-purchase sales are given as a percentage of the total sales for each of the past six months (they are not available for January and February), that the proportion of credit transactions has dropped considerably.

		Sound	Radiograms	Television)
January		93,000	35.000	103.000
February		99,000	33,000	98,000
March		95,000 (41%)	24,000 (62%)	85,000 (59%)
April		79,000 (42%)	I 6,000 (63%)	75,000 (59%)
May		75,000 (43%)	15,000 (68%)	62,000 (59%)
June		74,000 (40%)	13,000 (67%)	58.000 (57%)
July		84,000 (41%)	13,000 (67%)	61,000 (61%)
August	•••	73.000 (33%)	11,000 (54%)	64,000 (53%)
		677,000	160,000	606,000

Engineers in the B.B.C.

OPPORTUNITIES for graduate engineers and physicists in the B.B.C. are outlined in a booklet "Engineers in the B.B.C." Of the 4,500 employed in the engineering division, 600 are fully qualified engineers and physicists and about 2,300 "engineers who may not necessarily be fully qualified academically" who are mainly employed on operations and maintenance.

Each year the B.B.C. offers a number of two-year apprenticeships to graduate engineers and physicists who obtain Honours degrees. The course includes a period of training at the B.B.C. Engineering Training Department at Evesham, six months' practical works training in industry and some months in two or more of the specialist departments of the B.B.C.

The booklet can be obtained from the Engineering Establishment Officer, B.B.C., London, W.1.

Welsh V.H.F. Service

PENDING the construction of permanent v.h.f. stations in North Wales, a temporary v.h.f. transmitter has been installed at the Penmon, Anglesey, mediumwave station and was brought into service on October 2nd. The transmitter, which is horizontally polarized, radiates the Welsh Home Service on 94 Mc/s with a power of about 1 kW. The station will eventually have three transmitters—one for each programme.

V.h.f. transmitters are already being installed at Wenvoe, Glam., which is is hoped to bring into service by the end of this year.

PERSONALITIES

F. S. Barton, M.A., B.Sc., F.Inst.P., M.I.E.E., has left the Ministry of Supply, where he has been since 1946 (latterly as principal director of electronics research and development). to become defence supply adviser to the High Commissioner in Ottawa. In 1922 he joined the radio department of the Royal Aircraft Establishment, Farnborough, of which he became deputy head in 1936, and from 1941 to 1946 he was director of radio engineering with the British Air Commission in



F. S. BARTON.

Washington. His father, Professor E. H. Barton, F.R.S., D.Sc., was a pupil of Hertz in Bonn.

D. H. Black, M.Sc., Ph.D., F.Inst.P., who has been in the Civil Service since 1939, succeeds Mr. Barton as principal director of electronics research and development at the Ministry of Supply. For two years from 1943 he was on loan to the Admiralty after which he spent two years at R.R.D.E. (now R.R.E.), Malvern, before being appointed director of electronics research and development (defence) at the M.o.S., a post which he held until 1953 when he went to Australia as head of the United Kingdom Ministry of Supply staff in the Commonwealth. Before joining the Civil Service he was 14 years in industry, spending the last seven with Standard Telephones and Cables. Dr. Black, who is 56, is succeeded in Australia by Dr. W. H. Wheeler.

Dr. A. H. M. Arnold is leaving the National Physical Laboratory, where he has been head of the electronics section of the electricity division, to become reader in electrical engineering at King's College, London. Dr. Arnold, who in 1923 graduated from Liverpool University from which he received the degrees of Ph.D. and D.Eng., spent a year at Metropolitan-Vickers before going to the N.P.L. in 1926. He is 54.

F. H. Townsend, M.I.E.E., director and general manager of Cathodeon, Limited, Cambridge, was elected vice-president of the British Amateur Television Club at the second Amateur Television Convention recently held in London. Mr. Townsend ioined the Pye Group seventeen years ago and has been with Cathodeon (a member of the Group) since its formation during the war.

J. D. Parker, B.Sc.(Hons.), is the new secretary-general of the Comité International Radio-Maritime, of which H. C. Van de Velde is president. Colonel Parker, who is on the Regular Army Reserve of Officers, was in Royal Signals during the war and from 1941 to 1945 was in charge of the planning and technical section of the interservices research bureau. After the war he joined the Control Commission in Germany as controller of radio. Before being called up in 1939 he was in the G.P.O. W. A. Roberts, A.M.I.E.E., has been appointed regional engineer (Midland Region) by the B.B.C. in succession to J. A. Cooper, B.Sc.(Eng.), A.M.I.E.E., who is retiring after 32 years' service with the Corporation. Mr. Roberts, who joined the engineering division of the Corporation in 1937, became assistant to the chief engineer in 1949. Two years ago he was seconded to the Colonial Office as technical adviser on broadcasting development in the Colonies and was a member of the Commissions appointed to study broadcasting in the Gold Coast and Kenya. He returned to the B.B.C. a few months ago. Mr. Cooper, who has held his present position since 1950, was previously engineer-in-charge at Birmingham.

Michael Cooper, B.Sc., Grad.Inst.P., has been appointed planning engineer for teleciné and telerecording by Granada TV Network, Limited, the Monday-to-Friday programme contractors for the I.T.A. Lancashire station. He had been with the B.B.C. since 1951 where he was concerned with the planning and installation of studio and teleciné equipment. He was previously with Pye (1945 to 1946) and the G.E.C. Research Laboratories (1946 to 1951).

Granada TV Network, Limited, have also appointed John C. McKenzie, B.Sc., as television engineer for studio planning. In 1952, on leaving the University of Southampton where he obtained his degree, he was appointed research engineer at E.M.I. Research Laboratories where he was engaged on the initial experiments with the completely stabilized c.p.s. Emitron camera tube. For the past year he has been in the B.B.C. planning and installation department.

W. C. Pafford, who is both a television engineer and a lighting engineer, has left the B.B.C. after 23 years' service to join the Associated Broadcasting Company (now Associated Television, Ltd.), programme contractors for the London (week-ends) and Midland (week-days) I.T.A stations. He became a maintenance engineer at Alexandra Palace in 1936 and was later in charge of maintenance and war-time operations at the station.

L. E. C. Hughes, Ph.D., M.I.E.E., who was for some years with Londex, Limited, has recently been appointed head of the technical publications department of Ultra Electric, Limited. We understand that he is compiling a reference book on electronics applied to industry.

Walter Titmus, Assoc.Brit.I.R.E., has taken charge of the recently formed industrial controls division of the Solartron Electronic Group, Limited, which he joined in June after six years with Plessey's. From 1929 to 1938 he was with the Marconi International Marine Communications Company. In 1938 he joined the Army, becoming Captain (R.E.M.E.) in charge of the First Airborne Division's mobile and base radio workshops. From 1945 until he joined Plessey's in 1949 he was with E. K. Cole.

H. S. Payman, B.Sc., M.I.E.E., has been appointed general manager of Suflex, Limited, who are now manufacturing polystyrene capacitors in addition to their normal range of sleeving, wire and braided products. He was previously deputy director of component production and general manager of A.B. Metal Products.

F. Szekely, B.Sc.(Eng.), Grad.I.E.E., formerly on the staff of the B.T-H. Co., has joined the radio division of the Edison Swan Electric Co., Ltd., to promote the application of semi-conductors. After completing his graduate apprenticeship with B.T-H. he went into the research laboratory.

OUR AUTHORS

S. E. Gent, who, with D. J. S. Westwood, writes in this issue on vision a.g.c., has been with Ferguson Radio for seven years and for the past two years has been in charge of one of the Company's television laboratories. He was previously with Peto Scott where, during the war, he was working on military projects and subsequently on domestic radio equipment. He graduated

from the Northampton Polytechnic in 1949. Mr. Westwood joined Ferguson in 1953, having previously been with Cossor. He obtained his engineering degree in 1949.

W. Woods-Hill, author of "Electronic Digital Computors," is manager of research and development in the Calculator Research Laboratory of the British Tabulating Machine Co., Ltd. During the war he spent several years in radio and radar development units of the R.A.F., and subsequently became liaison officer attached to the French Air Force during the installation of their radar school at Auxerre, Yonne. He operates an amateur transmitter with the call G4JV.

A. L. P. Milwright, who describes a new microwave course beacon in this issue, is a principal scientific officer in the Royal Naval Scientific Service. He is at the Admiralty Signal and Radar Establishment, Portsmouth, and is in charge of the group which carries out development on civil marine navigational aids for the Ministry of Transport and Civil Aviation.

IN BRIEF

Receiving Licences.—An increase of 60,832 television licences during August brought the total in force in the United Kingdom at the end of the month to 4,786,415. These, together with the 283,473 for sets fitted in cars and the 9,054,699 for sound only, made an overall total of 14,124,587.

Vice-Admiral Dorling, director of the Radio Industry Council, will open this year's Amateur Radio Exhibition, organized by the Radio Society of Great Britain, at 12 noon on November 23rd, at the Royal Hotel, Woburn Place, London, W.C.1. It will close on the 26th.

R.S.G.B. News Service.—British amateurs now have an official broadcast news bulletin which is radiated on a frequency around 3,600 kc/s on Sunday mornings from 11.0 to 11.20 G.M.T. The bulletin is broadcast first on R/T, then keyed in abridged form and repeated on R/T. The station, call GB2RS, is operated at Walton-on-Thames, Surrey, by F. Hicks-Arnold (G6MB).

Science and Management.—The programme of the National Conference of the British Institute of Management, which will be held at Harrogate on November 2nd to 4th, includes a number of papers of interest to electronics engineers. Dr. B. V. Bowden, principal of the College of Technology, Manchester, will speak on electronic processing of data for management and John Diebold (U.S.A.) will deal with automation. Details of the Conference, the theme of which is "The impact of science on management in the future," are obtainable from the Institute, Management House, 8, Hill Street, London, W.1.

The 100 papers (ranging from transistors to transducers and communication theory to circuit design) presented at the 11th National Electronics Conference held in Chicago at the beginning of October will be published in volume 11 of "Proceedings of the National Electronics Conference." It will be available early next year, price \$5 from 84, East Randolph Street, Chicago 1, Illinois.

A steady growth in the membership of the **Tensor Club** of Great Britain during its first five years is recorded in the report sent to us by the joint organizers, Dr. W. J. Gibbs and S. A. Stigant. It now has a membership of 137 with over 50 per cent outside the U.K. Gabriel Kron, of America, an authority on tensor analysis, is patron of the club and during a recent visit to this country gave a series of lectures on the subject in London. Details of the club are obtainable from S. A. Stigant, 7, Courtlands Avenue, Hayes, Kent.

The annual dinner of the **Telecommunication Engineering and Manufacturing Association**, of which H. Faulkner is director, will be held at the May Fair Hotel, Berkeley Street, London, W.1, on November 10th. The Association is concerned mainly with general policy matters in the telecommunication manufacturing industry. **Television Test Signals.**—At any time between 10 a.m. and 11 p.m. (Monday to Friday) or 9.30 a.m. to 11 p.m. (Saturday) when programmes are not being radiated from the I.T.A. Croydon transmitter, a test signal on full power is now transmitted. No test signals are radiated on Sunday. The B.B.C. test period is from 10 a.m. to 1 p.m. (Monday to Saturday) and for the usual few minutes before the opening of each programme. The times of the test transmissions from the Belling & Lee transmitter (G9AED) at the I.T.A. site at Lichfield have been modified slightly. They are now 10 a.m. to 1 p.m., 3.0 to 6.0, and 7.30 to 8.30 (Monday to Friday) and 10 a.m. to 1 p.m.

R.I.C. Dinner.—Sir Walter Monckton, Q.C., Minister of Labour and National Service, is to be the guest of honour at the annual dinner of the Radio Industry Council at the Dorchester Hotel, Park Lane, London, W.1, on November 23rd.

The Second Congress of the International Commission on Acoustics will be held from June 17th to 24th, 1956, in Cambridge, Massachusetts, in conjunction with the fifty-first meeting of the Acoustical Society of America. The secretary of the Congress is John A. Kessler. Acoustics Laboratory, Massachusetts Institute of Technology, Cambridge 39, Mass., U.S.A.

Films on Loan.—Two more electronics films have been added to the list of those available on free loan from the Central Film Library of the Central Office of Information. They are "Ultrasonics" (20 minutes) and "Manufacture of radio valves" (26 minutes), both produced by Mullard. Particulars are obtainable from the Central Film Library, Government Building, Bromyard Avenue, Acton, London, W.3.

Demonstrations of Closed-Circuit Television are being given by a team of R.C.A. engineers at three international trade fairs being held in the Far East during the closing months of this year. The fairs are in India, Indonesia and Pakistan.

Electron Telescope.—With the object of showing the possible application of electronics to astronomical observations Pye fitted television equipment to the telescopes at Dublin Observatory and gave demonstrations to delegates attending the ninth general assembly of the International Astronomical Union held in the city in September. The meetings were attended by 500 delegates from 34 countries.

A course of six lectures on Magnetic Amplifiers is to be held on Wednesdays at 7.0 at the Technical College, Bradford, commencing on October 19th. (Fee 30s.)

Interference from Television Receivers.—A decree laying down the permitted limits of interference from television receivers in the broadcasting bands between 150 kc/s and 30 Mc/s was recently made in France.

A new 20-kW short-wave broadcasting station has been brought into service by the **Tanganyika Broadcasting Service.** It replaces an ex-Army 250-watt transmitter which has been in use since 1951. The transmitter, studio equipment and aerial system were supplied by Marconi's.

"Long-Range Television Reception."—In this article in the October issue, the heading for two of the screen pictures on p. 505 was unfortunately lost in the course of printing. It should have been Switzerland.

Since the Classified Advertisements pages were passed for press, it has been found that an incorrect advertisement for **Shirley Laboratories**, **Limited**, appears on page 159. The correct prices for the equipment are those given in Shirley's first advertisement under "New receivers and amplifiers."

The transmitting equipment for the I.T.A. Midland Station is being supplied by Pye and not to, as stated in the note on page 496 of our last issue. Since the publication eighteen months ago of the U.N.E.S.C.O. book "Television: A World Survey," there has been a large increase in the number of television stations in operation and many more countries have made known their plans for the future. A supplement to this international survey has therefore been published (H.M.S.O. 3s.) bringing up to date the information on the television service in 58 countries and non-self-governing territories.

The annual Electrical Industries Ball in aid of the funds of the Electrical Industries Benevolent Association will be held at Grosvenor House, Park Lane, London, W.1, on November 11th. Tickets, price $2\frac{1}{2}$ guineas, are obtainable from the E.I.B.A., 32, Old Burlington Street, London, W.1.

The first three Scholarship Awards, under the scheme announced a few months ago, have been awarded by the General Electric Company. These scholarships are to be awarded annually to selected members of the G.E.C. staff and provide for degree courses, postgraduate research or specialized studies.

BUSINESS NOTES

A contributing factor in the recent successful trials of the **Pye** underwater television camera on Lake Zurich, Switzerland, was the combined lifting rope and cable produced by **B. I. Callender's Cables.** It is a double-plasticsheathed version of a standard polythene-insulated television camera cable with a hemp covering; overall diameter is $1\frac{1}{2}$ in. By using this combined cable the time for lowering the camera, which weighs $2\frac{1}{2}$ cwt, was reduced from 45 to two minutes.

Lecture demonstrations of high-quality sound reproduction have been given by Goodmans Industries, Limited, in some fifty towns throughout the country during this year. The next and probably the last town to be visited in the present series is Luton, where, on November 15th and 16th, a demonstration will be given at 8.0 p.m. in the Town Hall. Tickets are obtainable from S. Farmer and Company, of Luton.

In the report on the S.B.A.C. Exhibition in our last issue we referred, on page 492, to the **Burndept** u.h.f. aircraft receiver (Type BE.234) as employing a frequency multiplication of 12 to obtain the spot operating frequency. The multiplication figure should have been six.

"Carcinotron."—Our attention has been called to the fact that the word "Carcinotron," used in our June issue as a general term to describe backward-wave oscillator valves, is the registered trade mark of the Compagnie Générale de Télégraphie sans Fil.

The installation of Pye v.h.f. radio-telephone equipment in the vessels of two more Thames lighterage companies brings the total number of users of this equipment on the London river to 32 companies operating over 300 sets. **Pye Marine** have also recently installed a station at the Dock Master's office in the port of Aberdeen and equipment in a pilot cutter and a dredger operated from the port.

The air traffic control and communications equipment at the re-opened Lympne Airport, Kent, is being provided by International Aeradio, Limited.

A mobile radio-telephone-equipped office has been put into service by the Automobile Association. Constructed by British Films, Ltd., of Balham, London, on an articulated semi-trailer, it is equipped with Pye R/T gear. A feature of the unit is the hydraulic 60-foot telescopic aerial mast built into the rear of the vehicle.

Radio communication equipment and radar navigational aids have been ordered from Marconi's by Shell Tankers, Limited, for eight tankers now being built. Marconi's have also recently provided radio and radar gear for new trawlers for Fleetwood and Grimsby.

The sales and order departments of McMichael Radio, Limited, are now at Wexham Road, Slough, Bucks. (Tel.: Slough 24541). Taylor Electrical Instruments, Limited, of 419/424, Montrose Avenue, Slough, Bucks, are offering service technicians allowances varying from £2 5s to £10 5s on Taylor instruments (up to about 14 years old and in working order) taken in part exchange for one of their latest models. The reconditioned models will be offered at a very low price to amateurs and students.

Ex-Government anti-aircraft radar mounted on trailers (Type AA3, Mark II) have been reconditioned by Winston Electronics, Limited, of Shepperton, Middx, for Government meteorological work, especially in the tropics. The reconditioning includes the substitution of a spark-gap modulator for the thyratron system.

E.A.P. (Tape Recorders), Limited, have moved from 546, Kingsland Road, London, E.8, to larger factory premises at 9, Field Place, St. John Street, London, E.C.1 (Tel.: Terminus 9627).

OVERSEAS TRADE

Canada.—A million-dollar contract for the construction of a mobile flight and tactical simulator to familiarize aircrews of the Royal Canadian Navy with the Grumman S2F anti-submarine aircraft, has been placed with Redifon, Limited. It combines facilities for flight familiarization with training in the use of radar and tactical anti-submarine equipment.

Component Exports.—According to figures issued by the Radio and Electronic Component Manufacturers' Federation, one-half of the present output of component manufacturers in this country, who now turn out about four million parts each working day, is exported either direct or assembled in equipment.

Thirty-one manufacturers of components and accessories participated in the joint exhibit organized by the R.E.C.M.F. at the British Exhibition in Copenhagen, which closed on October 16th after an 18-day run. In addition to the combined component display (the stands for which occupied a floor space of 1,300 square feet) Ekco and Mullard had individual stands. So far this year, exports of components to Denmark are well below the figure for last year, when the record total of £205,000 worth was imported from this country.

A mobile demonstration unit has been brought into service by the Automatic Coil Winder & Electrical Equipment Co., Ltd., for the display and demonstration of test and measuring equipment. It is now on a Continental Tour covering Belgium, Holland, Germany, Denmark and possibly Sweden.

RADAR SCANNERS, d.f. loop and a mast carrying a variety of transmitting and receiving aerials are mounted on the flat roof of the new Marconi House, Melbourne Street, Newcastle-on-Tyne. The debot has a test and repair shop, battery charging room, stores department and offices. In the building are also the Newcastle offices of Marconi Instruments and the English Electric Company.



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The Ratio Detector

How It Works

HE frequency modulation receiver must perform two functions additional to those required of its amplitude modulation counterpart. These two functions are:

1. The suppression of any amplitude modulation of the f.m. carrier due to noise, interference or variation of response over the receiver passband and

2. The conversion of the frequency deviation of the carrier to an amplitude variation directly proportional to the frequency deviation.

There are many methods of suppressing amplitude modulation: diodes so biased that their loading effect



Below : Fig. 2. The phase discriminator.



By K. R. STURLEY,* Ph.D., B.Sc., M.I.E.E.

on the circuit varies with the applied signal: a saturated amplifier having an amplification factor inversely proportional to the amplitude of the input signal: waveform slicers that square the waveshape. The most satisfactory type of amplitude limiter is the saturated amplifier, whose output is constant and independent of the input after a certain value of input has been exceeded. Its main disadvantage is that it introduces attenuation because its output is generally less than its input and it adds to the cost of the receiver.

The great value of the ratio detector is that it combines the role of amplitude limiter with that of

frequency-to-amplitude converter, and at the same time gives appreciable gain. At its best it is not as good as the saturated amplifier followed by a frequency-to-amplitude converter, but a previous article1 has given the difference in signal-to-noise ratio as only 2dB. Seeley and Avins, who were the inventors of the ratio detector, describe how it works and indicate the salient design points, but their article² is not easy to follow. There have been some attempts since to suggest a simple explanation of its operation but none that the author has seen have been at all convincing and some have left "confusion worse confounded." This article hopes to remedy the position, so let us start with a statement of what the ratio detector is.

The ratio detector is a modified form of the phase discriminator with the diodes performing the dual role of detection and variable damping. It functions in the same way as the phase discriminator to convert the frequency-modulated input to a combined frequency- and amplitudemodulated signal. Its diodes are so arranged that they provide variable damping to any undesired amplitude modulation already present at the input, reducing overall gain when the signal amplitude tends to increase and increasing it when signal amplitude falls.

There are many possible variants of the ratio detector, but a common form is shown in Fig. 1. As with the phase

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Fig. 3. Output voltage/frequency curve for phase discriminator.

discriminator each diode obtains its driving voltage from the primary (common to both diodes) and from a half secondary. In this case the primary voltage is derived via a tightly coupled tertiary winding L₃ giving a voltage step-down of about 0.6. The reason for this is that otherwise diode damping is too great on the primary. The diodes are not connected back-toback as in the phase discriminator but are in series, with the cathode of one joined through a resistancecapacitance load to the anode of the other.

First let us discuss the action of the diodes in detecting the amplitude change of the f.m. signal. We can best approach an explanation via the normal phase discriminator shown in Fig. 2. The secondary of the coupled circuits is centre-tapped, and half its voltage in series with the primary is applied to a diode D1 and the other half also in series with the primary is applied to a second diode D2. The primary voltage is developed via capacitor C across the r.f. choke L₀ between the centre tap of the secondary and the centre point of the diode load resistors. The outputs of the two diodes are connected in phase opposition, and an overall output voltage/frequency curve similar to that of Fig. 3, is obtained when the primary and secondary are tuned to the carrier unmodulated frequency. By correct choice of coupling and damping of the circuits, a linear conversion from frequency to amplitude change is obtained be-tween the two limits YY'. The reason for the S shaped curve of Fig. 3 is best explained by reference to the vector diagram of Fig. 4. The primary voltage vector is V1 and the two half-secondary voltage vectors $\pm \frac{1}{2}V_2$ are shown in phase opposition because of the centre tap. At resonance the primary and secondary vectors are at right angles, but for other frequencies greater or less than resonance the two secondary vectors are tilted to positions such as, for example, $\pm \frac{1}{2}V_2'$ and $\pm \frac{1}{2}V_2''$ respectively. The amplitudes of the and $\pm \frac{1}{2}V_2^{\prime\prime}$ respectively. secondary vectors decrease as the off-tune frequency increases, and their loci are the circles shown. The primary voltage vector does not appreciably change its length until the off-tune frequency is large. Assuming V_1 remains constant, we see that the voltage VD_1 and VD_2 applied to the diodes have maximum values when the secondary vectors are at approximately 45°. This explains the turn-over points Z and Ζ'.

Let us now take the simplified phase discriminator circuit of Fig. 5(a), and see what happens to the detected voltages across the diode load resistors R1. When the carrier frequency is unmodulated and the

VDI 1/2 V." Voz

Fig. 4. Vector diagram of phase discriminator.





Fig. 5. (a) Simplified phase discriminator circuit. (b) Voltages in different parts of the circuit.

phase discriminator is correctly aligned, the magnitude of the r.f. voltage applied to D1 is the same as that applied to D2, so that the d.c. voltage across AB is equal and opposite to that across BC, giving zero voltage across AC. For example, V_{AB} might be + 3 volts and V_{BC} –3 volts. If the carrier frequency is increased to $(f_c + f_d)$, V_{AB} might rise to +4 volts and V_{BC} fall to -2 volts, giving a total change of 2 volts across AC. A sinusoidally frequency deviated carrier would produce the waveforms shown in Fig. 5(b). Though the d.c. voltages across AB and BC are of opposite polarity, the audio-frequency changes across these points are of the same sign and add to each other to give

 $\Delta V_{AB} = \Delta V_{BC}$ and $\Delta V_{AC} = \Delta V_{AB} + \Delta V_{BC} = 2\Delta V_{AB}$ The audio-frequency voltages across AB and BC will cause a.f. currents I_{af1} and I_{af2} to circulate in ABF and CBF as shown in Fig. 5(a). These two currents are equal and opposite in the centre limb and cancel each other. As far as the audio frequencies are concerned there is no voltage across FB and we could remove the link FB were it not that it provides the

d.c. return path for each diode through the primary generator E₁. The d.c. current from D1 cannot pass through D2 because D2 cannot conduct in the opposite direction. A point worth noting is that when the carrier is at f_c , and the phase discriminator is correctly aligned, there is no output if the carrier is amplitude modulated. We can easily see that this must be so because increase of $V_{\scriptscriptstyle AB}$ to +4 and $V_{\scriptscriptstyle BC}$ to -4 still gives $V_{AB} = 0$. If the carrier is not at f_c or the phase discriminator is not correctly aligned to f_c , amplitude modulation is not suppressed. Thus if the carrier is at $(f_c + f_d)$, $V_{AB} = +4$ and $V_{BC} = -2$ giving $V_{AC} = 2$: amplitude modulation increasing the carrier amplitude by 50% takes V_{AB} to +6 and V_{BC} to -3 giving V_{AC} = 3. That is why we find that when an f.m. receiver is tuned in, background noise, which is mainly amplitude modulation, is a minimum when the receiver is correctly tuned to the incoming carrier. It is one way in which one can ensure that tuning of an f.m. receiver is correct.

Let us now turn, to the ratio detector, whose simplified circuit is as shown in Fig. 6(a). Diode D2 has been reversed and the link FB removed. Removal of this link does not open-circuit the d.c. return path because D2 can now take the current from D1. Reversal of D2 has had another effect, for the d.c. voltage across AB must have the same polarity as that across BC, so that if $R_1 = R'_1$ and the r.f. voltages applied to the diodes are the same as for the phase discriminator.

 $V_{AB} = V_{BC} = 3$ volts. and $V_{AC} = V_{AB} + V_{BC} = 2V_{AB} = 6$ volts Since the two capacitances C_1 and C'_1 are equal they will also be charged to 3 volts each, and the voltage across FB will be zero.

Let us imagine that the carrier frequency increases to $(f_c + f_d)$. The r.f. voltage applied to C₁ increases to produce a d.c. voltage after detection of 4 volts whilst that across C'_1 decreases to 2 volts. The total voltage across A and C is unchanged at 6 volts, i.e., no a.f. voltage is developed across AC when the carrier is frequency-modulated. It is this fact which makes possible the use of the diodes for variable damping as well as detection. We could connect a large capacitor across the points AC without disturbing the detection of f.m. waves though we will appreciably affect the detection of amplitude modulation initially present on the f.m. wave. We shall see later that this is just what is required to give variable damping.

Since the voltage across AC remains at +6 volts, it follows that the voltage across BC is also constant at +3 volts. Now the voltage across FB is

 $V_{FB} = V_{FC} - V_{BC} = 2 - 3 = -1$ volt.

Conversely when the carrier frequency falls to $(f_c - f_d)$ the voltage of F rises to +1 volt above B. Thus when f_d is varying sinusoidally we have an a.f. voltage across FB and if we earth B we can take off an a.f. voltage from F through a suitable CR network to the a.f. amplifier valve. The a.f. coupling capacitance is necessary, for we must not disrupt the d.c. path by offering an alternative through E₃.

Provided the ratio detector is correctly aligned and balanced about B we find that like the phase discriminator it suppresses all ampli ude modulation when the carrier is at f_c . For example increase of carrier amplitude merely increases the voltages FC and BC together and their sum is always zero: the $C_1R_1C'_1$ \mathbf{R}'_1 is a balanced bridge circuit at f_c . At any other frequency the amplitude modulation is not suppressed. Failure to suppress amplitude modulation at f_e





Fig. 6. (a) Simplified cir-cuit of the ratio detector. (b) Voltages across different parts of the circuit.



Fig. 7. Tuned circuit shunted by a resistance load.

can occur due to out-of-balance. Thus if $R'_1 = 0.9R_1$ amplitude modulation suppression occurs at some carrier frequency $f_c + \Delta f$ giving a division of voltage across C_1 and C'_1 of 1 to 0.9.

The waveforms of the various voltages in the circuit when the carrier is frequency-modulated are shown in Fig. 6(b). A significant difference between the ratio detector and phase discriminator is that the a.f. output voltage, V_{FB} , is only half that, V_{AC} , available from the phase discriminator, i.e., there is a 6dB loss from the ratio detector connection. If this were all that resulted, the ratio detector would certainly not be popular. This disadvantage can, however, be offset because the series connection of the diodes and the fact that there is no a.c. voltage across the points AC allow the diodes to be used as a variable damping resistance without impairing their detection characteristics.

Before dealing with the changes necessary to allow the diodes to function in this manner, let us see what happens when a diode is shunted across an r.f. tuned circuit. In Fig. 7 a diode is shown in series with a resistance R across a tuned circuit. If the diode is a perfect unidirectional device of resistance R_d and there is no capacitance across R, not even strays, the diode conducts on each positive half wave and during this time reflects a resistance $R_d + R$ across the tuned circuit. Since this operates half the time it will be equivalent to $2(R_d + R)$ for the whole cycle. When R is shunted by a capacitance C and the time constant CR is much greater than $1/f_c$ but much less than $1/f_a(\max)$, where $f_a(\max)$ is the maximum audio frequency, two things happen to the r.f. current taken by the diode:---

(1) it will increase many times in value, and

(2) the time during which it lasts will be appreciably reduced.

The r.f. current increases because the capacitance

short-circuits R and its value is only limited by the voltage applied, the diode resistance R_d and the internal resistance of the r.f. voltage source. The time during which current occurs is reduced because the CR combination acts as a self-bias circuit preventing diode conduction until this self-bias voltage is exceeded. Current amplitude is increased to a greater extent than duration is shortened, so that the energy taken from the tuned circuit is greater. The result is that the diode presents a resistance R_E of about $\frac{1}{2}R$ to the tuned circuit.

With both types of circuit there is practically no change of resistance reflected from the diode with change of r.f. voltage amplitude, and this method of damping is of no use to us. Let us now take a diode with a CR circuit and apply a fixed bias voltage V of polarity to stop conduction (Fig. 8(a)). Until the r.f. voltage initially exceeds V no current flows and there is no resistance across the tuned circuit from the diode. When the r.f. voltage greatly exceeds V the action of the diode is almost independent of V and determined only by its CR circuit: it then reflects a resistance of $\frac{1}{2}R$ across the tuned circuit. At intermediate voltages the resistance is greater, so that if we plot a curve of equivalent resistance R_E from the diode against r.f. voltage we should get a relationship like that of Fig. 8(b). This does show the kind of variable damping characteristic we require, for resistance decreases as amplitude increases. If we operate with the tuned circuit in the anode of a pentode at an r.f. voltage of E_0' (Fig. 8(b)), any change of E_0 would produce a change of R_E and vary the gain of the pentode in a direction opposite to the input amplitude change causing it. As an example let us assume the following conditions

Turned circuit dynamic impedance, $\frac{L}{CR} = 0.5 M \Omega$

 g_m of value = 5mA/V

Equivalent resistance of diode at $E_0 = 9.8V$, $R_E = 40,000 \Omega$; $E_0 = 10V$, $R_E = 20,000 \Omega$; $E_0 = 10.2V$, $R_E = 13,000 \Omega$.

Fig. 8. (a) Tuned circuit shunted by a biased diode with RC load. (b) Resistance ($R_{\rm E}$) reflected from diode, plotted against r.f. voltage.



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The anode load of the pentode consists of $R_{\rm E}$ in parallel with 0.5M Ω , and the latter is large enough to be neglected. The gain of the valve at the different values of E_0 will be $g_m R_{\rm E}$, viz.,

E	Gain	Input Voltage
9.8	200	0.046
10.0	100	0.1
10.2	65	0.154

We see that, with the value we have chosen, an input voltage change equivalent to 50% amplitude modulation has been reduced to a 2% amplitude modulation of the output voltage. A still greater reduction of amplitude modulation could be obtained by a greater change of $R_{\rm E}$.

If a diode is to be used in a receiver as variable damping the fixed bias voltage can easily be obtained from the h.t. supply, but the effectiveness of the damping will be confined to a comparatively small range of output voltages about E_0' . Such a scheme would not be satisfactory in a receiver which was designed to accept either weak or strong signals and which had no additional limiting action before the diode damper. The ideal arrangement would be to have the bias voltage variable to suit the output voltage, decreasing bias when the output voltage is less than E_0' (Fig. 8(b)) and increasing bias when the output voltage is greater. This would allow us always to operate on the part of the $R_E - E_0$ curve giving the best amplitude suppression. This can be achieved by letting the diode provide its own self-bias, using an RC combination whose time constant is much longer than the reciprocal of the lowest amplitude modulation frequency likely to be encountered, i.e., about 0.1 sec corresponding to a frequency of 10 c/s. The circuit would be as shown in Fig. 9. C_1 and R_1 would be



Fig. 9. Variable damping diode.

about 300pF and 1000 Ω and C₂ and R₂ 25 μ F (electrolytic) and 10,000 Ω . Minimum damping resistance would be $\frac{1}{2}$ R₁ to any amplitude modulation change and we should have variable damping similar to that shown by the curve in Fig. 8(b). The operating output voltage E₀' will have a definite relationship to the bias voltage

V and will be approximately
$$rac{V_0}{0.9}=rac{R_1+R_2}{0.9R_2}$$
 V;

the 0.9 is included to allow for a diode detection efficiency (ratio of d.c. voltage to peak applied voltage) of 90%. If R_1 is made zero the output voltage operating point is almost coincident with V, the $R_E - E_0$ curve is very steep falling to almost zero damping resistance, and the diode suppresses (or considerably reduces) any positive amplitude change. For any negative amplitude change there is no damping from the diode because the bias V holds it non-conducting. Seeley and Avins refer to this condition as fully stabilized because all the detected d.c. voltage is used as the bias voltage V. It is obviously an undesirable condition in which to work and the more normal method is to operate with about 80% of the detected d.c. voltage stabilized, i.e., used as the bias. With this degree of stabilisation, the variable damping detector should appreciably reduce input voltage amplitude modulation of as much as 90% in either positive or negative direction. Naturally the amount of suppression in the negative direction will be determined by the undamped Q of the secondary and also the primary coil, since maximum gain cannot exceed the undamped gain. Q_2 and Q_1 should therefore be as high as possible.

So far we have discussed the effect of diode damping to any change of carrier amplitude, but it is also essential to know what the damping will be to a constant carrier amplitude because this will determine the gain of the stage to a frequency-modulated signal. The equivalent resistance reflected into the tuned circuit from the diode in Fig. 9 at a constant carrier amplitude is determined not by R1 alone but by the sum of $R_1 + R_2$, and it will in fact be nearly $\frac{1}{2}(R_1 + R_2)$ R.,). This is very important because it means that the gain of the stage to a frequency-modulated constant-amplitude signal can be reasonably large and the heavy damping is only effective on carrier amplitude changes. We are in fact making use of a low ratio of a.c./d.c. load in the diode circuit: only the d.c. load affects the carrier amplification whereas the a.c. load damps any changes in carrier amplitude.

Having explained the principle of variable damping by diodes let us now see how it is applied in the ratio detector to provide amplitude limiting. Let us imagine that we have a centre-tapped tuned circuit across which we wish to have variable damping. Clearly



Fig. 10. (a) Variable damping diode across a centre-tapped circuit and (b) in a balanced circuit.



Fig. 11. Alternative ratio detector connection.

we could place across one half of the circuit the variable biased diode as shown in Fig. 10(a). The r.f. bypass capacitance C₁ is shown connected direct to earth instead of to the junction of R1 and R2 as in Fig. 9. This has no effect on the operation of the circuit, but it does ensure that the r.f. performance is not influenced by possible inductance in the C, leads. Since the centre-tapped circuit is balanced it is preferable to have a balanced damping circuit and this can be realized by using two diodes connected in series across the whole tuned circuit with their common load resistance centre-tapped to earth, as in Fig. 10(b). It is only one step from Fig. 10(b) to the ratio detector of Fig. 1. The link from the centre of the coil to the junction of capacitors C_1 is replaced by the tertiary coil. We have already shown that frequency modulation of the carrier produces an a.f. signal across the two capacitances C_1 in Fig. 10(b) and that the resistance arm $(2(R_1 + R_2)$ carries only a d.c. current and no audio frequency. It is therefore possible to connect a large capacitor across a whole or part of the resistance without affecting the performance of the circuit as a detector of frequency modulation.

We have seen above that the large capacitance must not be used across the whole of the resistance load, otherwise no suppression of negative amplitude modulation occurs. Even if the negative modulation problem did not exist we could not use a large capacitance across all the resistance because if there were a quick reduction in amplitude of the f.m. carrier the bias could not fall fast enough and the diodes would be cut off. We should then have either no audiofrequency output from the f.m. carrier or else a very distorted output.

Suitable values for the components in Fig. 1 are $C_1=330 p F;$ $C_2=16$ to $25 \mu F$ (electrolytic), $R_1=1000\,\Omega,$ $R_2=6800\,\Omega.$ R_3 and C_3 are selected to give the required de-emphasis of the audio frequencies, possible values are 50,000 Ω and 1000pF and C₄ $(0.01\,\mu\text{F})$ is the audio-frequency coupling capacitance. With the component values quoted above the damping resistance effective at constant carrier amplitude will be approximately $(R_1 + R_2) = 7800 \Omega$ across the complete secondary and $\frac{1}{4}(R_1 + R_2) = 1950 \Omega$ across the tertiary winding. The reason for the lower damping resistance across the tertiary is that the r.f. currents from each diode traverse the tertiary coil. Since the tertiary winding has a step-down of about 0.6, the resistance reflected across the primary will be $1/(0.6)^2$ of that across the tertiary, i.e., it will be about the same as that across the secondary. This explains why a tertiary is essential in order to reduce the damping on the primary and so increase overall gain to the f.m. signal. Any amplitude modulation of the f.m. carrier will see an effective variable resistance whose minimum value is $\mathbf{R}_1 = 1000 \,\Omega$ across the secondary and $250 \,\Omega$ across the tertiary.

Many factors contribute to optimum amplitude limiting and not least the design of the coupled circuits, the undamped Q of which (particularly the secondary) should be as high as possible if effective suppression of negative amplitude modulation is to be achieved. A suitable design has already been given in this journal³. It is important, too, to operate the circuit in the balanced condition at the unmodulated carrier frequency, f_e . This is achieved by adjustment of one of the R_1 resistances until a.f. output is a minimum when an amplitude-modulated input carrier of frequency f_e is applied. The resistances R_2 should have tolerances of $\pm 1\%$. Amplitude suppression is, of

course, obtained from diode damping when the circuit is not balanced, but the suppression will be less effective when the carrier deviation, f_d , is positive than when it is negative, or vice versa.

Since the ratio detector offers a fixed damping resistance to any steady value of carrier amplitude it will be clear that a "semi-permanent" change in carrier amplitude must produce a corresponding change of audio-frequency output from the frequency-modulated carrier. Hence tuning from a strong to a weak incoming signal will cause the a.f. output to fall. With the saturated amplifier limiter there is no change in a.f. volume and we have perfect automatic gain control (a.g.c.) characteristics.

This disadvantage of the ratio detector can be partially overcome by using the d.c. voltage developed by the ratio detector diodes to provide an a.g.c. voltage to bias preceding i.f. amplifier stages. The The point from which a.g.c. could be obtained is indicated in Fig. 1. An alternative ratio detector circuit is shown in Fig. 11: the main difference between it and Fig. 1 is that the resistances R_1 are placed in series with C_2 which is split into two capacitors to maintain circuit balance. Adjustment of the earth tapping point on R_1 gives fine control of balance.

To sum up, we see that the success of the ratio detector is due to the fact that the resistance load of the diodes carries no audio frequency when the carrier is frequency-modulated. This allows us to use a low ratio of a.c. to d.c. load and by this means suppress amplitude change on the carrier.

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THE EDITOR LETTERS TO

The Editor does not necessarily endorse the opinions expressed by his correspondents

Fourier-Fact or Fiction?

IN his article on Fourier analysis in the September Wireless World, "Cathode Ray" becomes confused because of careless use of the term "frequency

He gives a simple and valid definition of frequency; viz., the rate at which identical events occur. It follows that any waveform can have one and only one frequency and it is meaningless to ask whether a waveform having only one frequency is necessarily sine. It is not necessarily anything.

Fourier's theorem states that any periodic function can be made up from a number of sine functions of different frequencies. It is common usage to say that a waveform contains frequencies other than its own, when what is meant is that the waveform can be made up from sine waves of these other frequencies, together with a sine wave of the fundamental frequency. The difficulty arises when "Cathode Ray" uses the word frequency without referring it to any particular waveform.

For instance, in asking what frequencies a sine waveform may contain, we are really asking what sine waves. with their harmonics, are needed to make it up. The answer is obviously the pure fundamental, which is where we came in.

Southampton.

G. F. FOOKS.

"CATHODE RAY" wishes, on p. 456 of the September issue, that some mathematician would chip in with a neat proof to support his arguments; but unfortunately the theorem he was hoping for is false. I have sent for his criticism what I think is a fairly neat proof that any waveform (within reason) together with the same one moved $\frac{1}{4}$ the way along the wavelength (the same relationship as the sine to the cosine) can be used at frequencies 1, 2, 3 \dots times that of another given wave function, to analyse the latter in "components" of the former.

So "Cathode Ray" is right to proceed to look for other features of the sine in order to justify its preference, and I admire the reasons he gives. But he overlooks one important one; namely, that with sines and cosines, the coefficients in the Fourier series (i.e., the amplitudes of the harmonics) are relatively easy to compute; if we had used any other basic waveform they would have been very much more difficult.

Braunstone, Leicester.

J. P. DOUGHERTY.

"Cathode Ray" replies: -

It is hardly surprising that my metaphysical musings

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on Fourier in the September issue aroused some comment, not to say protests.

Some correspondents have shown themselves still unable to look at the sine waveform with an open mind. G. F. Fooks scores a definite point, I think, in his criticisms. But even he lapses into sine worship in his last paragraph. My whole point was that, however advantageous it is in practice to regard sine waves as basic, in principle one could choose otherwise.

J. P. Dougherty has supported me in this with a mathematical proof, which looks all right, though I am not mathematician enough to pronounce with authority on its validity. Incidentally, what he seems to have taken for an appeal to establish the sine wave as the only possible basis of a "Fourier" series was intended to be a challenge. His proof that other waveforms could be used relieves me of the fear that someone would successfully accept the challenge, and I am much obliged to him for his mathematical blessing on what might seem my rather presumptuous questioning of the divine right of sines.

With regard to his second point, I thought I had at least implied it in the words (p. 457) "So if electrical engineering (for example) were based on any other waveform, the whole thing would be unbearably complicated and difficult." But it is just as well to have it stated plainly.

I am indebted to J. F. Stray for referring me to Edding-ton's "Philosophy of Physical Science," Chap. VII, in which Sir Arthur contends that some so-called laws of nature are in fact man-made. He likens these to a "law" a sculptor might propound—that shapeless blocks of marble naturally contain the form of a human head, his evidence being that such a form can be isolated by means of an experimental analysis using a hammer and chisel. What I would like to ask the distinguished author is what view he would take of the matter if the chips were all found to be of identically the same human form, and of sizes that were exact integral sub-multiples of the main head. "CATHODE RAY." head.

Neon Tuning Indicator

IN the article by John D. Collinson on page 429 of the September issue it is stated that "With finite values of μ and R_k the anode current changes will not be equal and opposite, but unbalance is immaterial as it reduces to zero at the working point." The second part of the statement is surely incorrect.

The changes in anode current do not matter provided the anode currents passed by each half of the valve at

the working points are equal and that the mean characteristics are equal. In short, the two halves of the circuit must be balanced statically at the working point.

In practice this seems very unlikely, especially as no selection between valves is possible since these will usually be in the same envelope. This is easily confirmed by shorting the input to the indicator and it will be seen that the neons do not glow equally except in exceptional choice of components. This means that the indicator will not indicate zero voltage for equal neon brightness but will indicate a constant error voltage, the error depending on the static unbalance of the circuit. This difficulty may, of course, be overcome by inserting

a small resistance in the cathode lead of one-half of the valve, thus altering the standing current balance and making the neons glow equally for zero input voltage.

The statement that the indicator indicates balance to within $\pm 3 \text{ kc/s}$ thus seems unjustified.

I should add that I am using the modified form of this indicator with a ratio detector and find it gives excellent results.

North Harrow, Middx. A. LIONEL WILLIAMS.

The author of the article writes: I agree that the unbalance only reduces to zero at the working point with perfectly balanced valves. The suggested modification is therefore valuable if it is thought that errors due to normal valve unbalance are serious.

In practice, however, it seems that commercial double triodes balance rather better than Mr. Williams implies. In the circuit given, a 20 per cent unbalance between anode currents requires 75 mV to bring them into balance. With a discriminator output of 100 mV per kc/s deviation this corresponds to a tuning error of ± 750 c/s. The figure of ± 3 kc/s quoted actually includes the ex-

pected valve error together with the human error in estimating the relative brilliance of the neons, the latter error being the greater. JOHN D. COLLINSON.

" Etched Foil Printed Circuits"

H. G. MANFIELD is to be congratulated on the very interesting and informative article in your September issue.

However, neither with nor without the benefit of Crown Copyright does copper liberate hydrogen from nitric acid and it is suggested that the word "gas" be substituted for "hydrogen" in an otherwise outstanding article. Hornchurch, Essex. L. D. STUAR

L. D. STUART.

Television Aerial Feeders

"DIALLIST'S" comments on balanced versus unbalanced television aerial feeder cables in your August issue conflict with my own experience so I cannot let them pass without some comment.

I have conducted reasonably careful tests using coaxial cable and both screened and unscreened twin cable and I concluded that:

(a) Feeder pick-up was only important when the feeder cable was in a strong interference field remote from the aerial (e.g., a long feeder cable in a building running very close to power supply cabling carrying a considerable level of interference signals, the aerial itself being mounted outside and well above the building).

(b) When there was appreciable feeder pick-up, the type of cable was unimportant but the preciseness of the match to the receiver was all important. A twin feeder natch to the receiver was an important. A twin receiver connected to an unbalanced input was as bad as a coaxial feeder connected to a balanced input. As "Diallist" will be aware, to achieve a good balanced input to a receiver is not all that easy. Even careful

attention to the design of the input circuit and care in positioning components will rarely produce a balance to-unbalance ratio at v.h.f. much better than 20 dB. Certainly a turn or so of wire over the grid coil of the first stage with both ends brought out to a second- or thirdrate socket cannot possibly give a balanced connection.

I have yet to meet a commercial TV set with a balanced input worthy of that description or with a twin aerial feeder socket really suitable for use at v.h.f. and I suspect that the cost of providing a really good balanced input is likely to make the most enthusiastic manufacturer jib. In these circumstances, since we are both agreed that good matching at the receiver input is essential, it would be better to recommend the set maker to stick to his unbalanced input and the set owner to install coaxial feeder cable.

Cheltenham, Glos.

T. A. LEWIS.

" Back to Methuselah"

I AM not sure what to make of "Free Grid's" reference to W. T. Cocking and myself as "irregular" contributors, coming as it does from such a source. However that may be, his crediting me with a mere 28 years' irregularity jerked me up out of my bathchair to seize reed pen and papyrus. If "Free Grid" will turn up the issue dated August 15th, 1923 (i.e., over 32 years ago), he will see

Adjust 15th, 1925 (i.e., over 52 years ago), ne will see on page 660 my contribution on a voltage raiser for valve transmitters. (Note particularly, *valve* transmitters.) On the other side of the page he may be interested in the photograph of an up-and-coming youngster, called Captain Eckersley, inspecting the new B.B.C. apparatus (installed in the studie) for relaying the London pro-(installed in the studio!) for relaying the London programme to transmitters in other parts of the country. I am sure "Free Grid" would find the wiring of this equipment a fit subject for one of his entertaining comments. Bromley, Kent. M. G. SCROGGIE.

Magnetic Tape

ALTHOUGH a number of theoretical claims have been made that tape recordings can be stored for extended periods, it would be interesting to know if any reader has stored a tape for five years and finds that its original quality has deteriorated by "print through" or other causes.

Pinner, Middx.

A. H. BEAN.



Driving a diesel farm tractor by radio control is hardly likely to become general practice, but it has been introduced by Ford's as a novel method of demonstrating their Fordson Major. There are six controls covering steering (L & R), clutch, implement lowering and raising, and engine switch. The radio equipment, which was designed by G. Honnest-Redlich, of Radio & Electronic Products, is operated on the tone-modulated carrier system, the carrier being 27.12 Mc/s and the modulating frequencies between 250 and 400 c/s.

Dual-band Television Aerials

1-Alternative Designs for Combined Band I and Band III Reception

By F. W. R. STAFFORD,* M.I.E.E.

bY now everybody concerned with television understands that Band I is the frequency spectrum embracing the five existing B.B.C. television channels and Band III is the v.h.f. spectrum which will ultimately be devoted, exclusively, to television and will be divided into eight channels. In round figures Band I covers from 41-68 Mc/s and Band III from 175-215 Mc/s. This brings one to the first important point, namely, the frequency ratio of Band III to Band I. The lowest value is about 2.6/1 and the highest is about 5.3/1.

Consider a simple vertical half-wave dipole resonant at 45 Mc/s (Fig. 1). For a vertically polarized wave arriving normal to the axis of the dipole (see the arrow) this aerial will develop an open-circuit e.m.f. across its terminals x, y, given by:

$$e_{o} = \frac{E\lambda}{\pi}$$
 (1)

where E is the incident field strength in volts/metre and λ is the wavelength of the radiation.

Notice that the vertical directional characteristic as given by the dotted line shows a maximum in the direction of arrival normal to the dipole axis.

In these conditions the dipole is equivalent to a generator possessing an internal resistance of approximately 80 ohms and if terminated by an 80-ohm load, conveniently through a transmission line of characteristic impedance Z_0 =80 ohms, the voltage across the load will be given by:

$$e_1 = \frac{E\lambda}{2\pi} \qquad \dots \qquad \dots \qquad \dots \qquad \dots \qquad (2)$$

assuming that the line losses can be neglected.

This is all well known but has to be reintroduced as the basis for considering what happens when one attempts to use the same aerial at frequencies other than the fundamental resonance, and particularly over the Band III spectrum.

Three effects must be taken into consideration because they all act in a way which reduces the e.m.f. at the terminals x, y, and which can never attain the value at half-wave resonance if loaded with 80 ohms. First, the induced e.m.f. becomes a most complicated function of frequency. Second, the impedance as seen at the terminals varies with frequency. It is high (several hundred ohms) when the frequency ratio is about an even integer and falls to about 80 ohms when the ratio is about an odd integer. Third, the vertical directional characteristics split up into a number of lobes and never show a principal maximum in the direction of arrival normal to the axis of the dipole, which is the direction which is used for all practical purposes. A particularly bad case is when the ratio is 4 to 1 in which case the induced e.m.f. is zero at normal incidence (Fig. 2). This would result in minimum signal reception and maximum interference from aircraft (flutter) and ignition interference. *Belling and Lee Ltd.

It is beyond the scope of the article to analyse these three effects because it involves laborious mathematics and seems to defy simple explanation. It is one of those problems which is best solved by experiment and measurement. Perhaps one of the reasons why no one has bothered much about it is because it would be bad engineering to attempt to use a resonant dipole at far-removed frequencies, but the television aerial installer should be in a position to know just how Band I aerials behave on Band III, particularly on certain channels. It so happens that the range of Band III/Band I ratios (2.6 to 5.3) provides all the conditions for adverse performance, from gross mismatch to the load impedance, to splitting up of the vertical directional characteristics.

Very careful measurements have been made of the relative gains of Band I aerials over the Band III spectrum and the results are shown in the family of curves of Fig. 3. A resonant dipole on each Band I channel was fed through a balanced 80-ohm transmission line to an 80-ohm load, and thence to an accurately calibrated measuring receiver. Spot measurements were taken at each of the eight channels on Band III, (a) with the Band I aerial under consideration, and (b) with an optimized resonant half-wave dipole on the Band III channel under consideration. The ratio of these two amplitudes expressed in dB





Fig. 3. Response of Band I dipoles (Channels 1-5) over Band III.



Fig. 4. Channel l or 2 dipole adapted to provide Channel 9 reception at moderate efficiency.



Fig. 5. Showing how the Band III vertical directivity is affected by the Band I elements.

gives the loss one incurs by attempting to use the Band I aerial on Band III.

Apart from this loss, which is particularly high for dipoles designed for Channels 1 and 2, the vertical directional characteristics take the form of lobes directed upwards and downwards thereby increasing the susceptibility of the installation to aircraft flutter in respect of the upward lobes, and ignition interference for the downward lobes (Fig. 2).

The performance of Band I multi-element aerials follows the same pattern over Band III as the simple dipole, because the various parasitic elements are so hopelessly out of resonance that their contributions can be ignored. It would take a great deal of work to check all the possible arrangements, but a few odd measurements support this view.

At the present time we are concerned with the dual reception of Channels 1 and 9 in London, 2 and 9 in Lancashire and 4 and 8 in the Midlands.

It will be seen from Fig. 3 that dipoles for Channels 1, 2 and 3 are extremely inefficient in the middle of Band III where the only television allocations to date exist. Not one of the five dipoles can be said to be moderately efficient over the whole of Band III, but a Channel 4 dipole is reasonably responsive on Channel 8. This observation will be dealt with later.

Notwithstanding the poor performance of Band I aerials on Band III, particularly Channels 1 and 2, there are so many of them that serious consideration has to be given to ways of improving their performance on Band III without introducing any serious loss on Band 1. The attachment of additional elements of appropriate dimensions can effect considerable improvement, but, according to the author's tests, cannot bring the aerial up to the performance of a simple half-wave dipole designed for the particular Band III channel which it is desired to receive. At best the Band III performance is about 3 dB below that of an optimized dipole, that is, with the longer elements wholly removed (Fig. 4(a)).

This method of adaptation is thus guite useful in areas of high Band-III field strengths, but is only usefully confined to Channel 1 (London) and Channel 2 (Holme Moss) dipoles. Both of these exhibit a high impedance at the terminals x, y, on the I.T.A. Channel 9, since the frequency ratio is about an even integer Thus we may consider the equivalent circuit (4/1).on Band III to be a generator with an internal resistance of 80 ohms whose output terminals x, y, (Fig. 4(b)) are shunted by an impedance (Z) of several hundred ohms. In view of the high shunt value of Z why should the performance be measurably inferior to a simple Band III dipole? The answer is, simply, that the vertical directional characteristics are modified because the Band I elements are contributing adversely with their upward and downward lobes. This is best seen by using the dipole as a radiator and transmitting energy from it. For a given energy applied to the terminals x, y, most of it is radiated in all directions at right angles to the axis of the dipole on Band III if the Band I elements are removed. On replacing these elements some of this energy is used for radiating upward and downward lobes which serve no useful purpose and must reduce the energy radiated in the desired direction. By the principles of reciprocity the directional characteristics of a given aerial are identical for transmission or reception which proves the point. Fig. 5 may assist the reader in following this explanation.

A better method of converting Channel 1 and 2



Fig. 6. Co-phased Channel 1-2, 9 conversion.

aerials to Channel 9 will now be described.[†] The Band I dipole is surrounded by a metal sleeve of approximately two inches diameter, symmetrically disposed about the junction (Fig. 6(a)) and is approximately half a wavelength from tip to tip on Channel 9. In practice the cylinder may be dispensed with and replaced by at least two metal rods suitably spaced off with insulators, although one rod does work fairly well. In this arrangement the sleeve (or rods) act as a continuation of the low-impedance coaxial transmission line normally connected to x, y so that at points Q and P, where the current is maximum and the impedance match. Of course, the current is maximum

at Q and P because they are approximately $\frac{3n}{4}$ from

the outermost tips on Channel 9, always remembering that we are discussing Channels 1 and 2 dipoles, where the frequency ratio is of the order of four to one. Moreover, the upper and lower elements are fed in phase so that this co-phased conversion should be almost as efficient as *two* resonant Channel 9 dipoles placed one above the other and co-phased. Then the increased gain would be 3 dB. The measured gain of this co-phased conversion is about 2 dB above a simple resonant dipole. A further very good property is seen in Fig. 7 which shows that the vertical directivity is concentrated along the normal direction of arrival of the signal, thereby reducing aircraft flutter and ignition interference to a greater degree than that possible by the former adaptation of Fig. 5.

So far the adaptations described do not confer any horizontal directivity on either aerial. Of course, the Band I aerial can have its usual parasitic reflector and/or directors added to give it horizontal directivity but the Band III performance will still be that of an omni-directional dipole.

Observations on the transmission of the experimental station G9AED (Channel 9) were conducted in a mobile research unit in which the vision signal (test card) could be compared on (a) a simple omnidirectional dipole and (b) a three-element directional array. In many instances serious "ghosts," due to delayed reflections, were noticeable when the dipole was in use, but in most cases they were eliminated by changing over to the three-element array. "Ghosts" were certainly far more prevalent on Band III than on Band I as was predicted in an earlier article by the author¹. Objects such as churches, large blocks of flats, and electricity cable pylons were often respon-

¹Patents pending. ¹F. R. W. Strafford, "Baud III Television Aerials." Wireless World, April 1954.



Fig. 7. Vertical directivity of co-phased conversion.



Fig. 8. An adapted Channel I-2, 9 aerial providing gain and horizontal directivity on Channel 9.

sible, while cooling towers and gasholders were outstandingly bad, and so were sharply rising hills, if they were sited behind the receiving installation. Thus while an adapted Band I dipole may not give rise to "ghosts" on that band there may be serious "ghosts" on Band III.

The next step is to design an adaptor which can provide horizontal directivity on Band III, and a neat way of achieving this for Channel 1 and 2 adaptations will now be described with reference to Fig. 8⁺.

As usual the feeder to the receiver is connected to the Band I dipole terminals x, y, and the folded dipole is correctly dimensional for the Band III channel required, i.e. the distance bc approximates to one half wavelength. Consider the conducting loop $x \ a \ b \ c \ d \ y$; this may be regarded as a section of transmission line short-circuited at the point P. From normal transmission line theory a short-circuited

line of length $\frac{n\lambda}{4}$ where *n* is odd, behaves as a high

impedance at its input terminals (in this case x, y). Therefore, by making the length $x \ a \ b \ P$ such that this condition obtains on Band I a high impedance is shunted across x, y which will absorb negligible energy, since it is in parallel with about 80 ohms, i.e. the radiation resistance of the Band I dipole, and will therefore not adversely affect its performance.

On Band III the Band I element presents a high impedance at the terminals x, y but the folded dipole which has a terminal impedance at a, d of about 300 ohms on Band III, is transformed down to 80 ohms at the junction x, y by the parallel conductors x, a and y, d by the familiar stub technique. The spacing, S, is chosen to satisfy the characteristic

impedance required for the parallel section to achieve the desired transformation from 300 to 80 ohms.

This system may now be made directional by the addition of one or more directors D_1 , D_2 , etc. The same result could be achieved by connecting the folded dipole terminals a, d to the plain dipole terminals x, y by means of a diplexer, which is a combined high-pass low-pass filter with inter-band cut-off frequencies,

but the use of a diplexer is confined mainly to other aspects of installations and will be discussed later. In this particular application, however, it is a more expensive way of achieving the same result.

As in the case of the simple adaptations of Figs. 5 and 6 the method can be extended to Band I dipoles possessing a reflector and/or directors.

(To be concluded)

New Vision A.G.C. System

Advantages Obtained from an Improved Method of Gating

By S. E. GENT,* B.Sc. (Eng.), Grad.I.E.E., and D. J. S. WESTWOOD,* B.Sc. (Eng.)

W HATEVER the argument for using vision a.g.c. in the past, there is little doubt that, with the advent of Band-III television and the introduction of more than one programme, an efficient system of this kind is now a highly desirable, if not essential, part of all except the very cheapest of receivers.

The main reason for this requirement is the certainty that in most districts the Band-I and Band-III signals supplied to the receiver will be of different strengths, and it will not be desirable for the viewer to have to readjust the gain control of the receiver when switching from one programme to the other. One might argue that this could be overcome by pre-set gain controls in the receiver, one for each channel, which could be adjusted by the service technician and then left. But even if this is accepted for two or even three channels it must be remembered that there will one day be more channels in use, and the prospect of thirteen pre-set gain controls is a little staggering, to say the least! It is to be expected that variations in signal strength on Band III may be more serious than on Band I, and this may be used as a further argument in favour of a.g.c.

The majority of current a.g.c. systems will not deal efficiently with aircraft flutter unless the flutter frequency is very low, but it is claimed that the new system described in this article can be better in this respect than previous systems.

Obviously, in order to deal adequately with com-



Fig. 1. Simple a.g.c. system using mean d.c. voltage at sync separator grid.



paratively fast changes of signal strength, such as are caused by aircraft flutter, the time constant in the a.g.c. smoothing circuit must be short, and the a.g.c. voltage must be derived entirely from the vision signal and must be independent of the picture information and the sound signal.

Also, in order to cater for all likely ratios of Band-I to Band-III signals, the a.g.c. should have a large range of control over the gain of the receiver, say 60dB, and there should be no necessity for a pre-set sensitivity control. With a large control ratio such as this, care must be taken to delay the control on the r.f. stages relative to that on the i.f. stages, and then to proportion it correctly, if a good signal-to-noise ratio is to be maintained over the whole control range of the system. This necessitates a larger a.g.c. voltage than if no delay were used.

Moreover, in order to simplify the viewer's task in adjusting his receiver, the a.g.c. operation should be independent of other adjustments, for example, time-base locking.

It was with these points in mind that this new system was devised, but before describing it we will give a brief survey of the other methods at present in use and their shortcomings.

Commonly Used Systems.—The simplest form of a.g.c. uses the mean d.c. voltage developed as a result of diode action at the sync separator grid. The general arrangement is shown in Fig. 1. The contrast control usually operates by applying a positive bias to the a.g.c. control line, thereby cancelling part of the developed negative bias. A diode is included, termed a delay diode, and this clamps the control line to earth for signals below a certain level which is determined by the setting of the contrast control.

There are three objections to this system. The smoothing circuit must deal with the 50-c/s video and sync components, resulting in rather a long time constant. On pictures of small d.c. level, such as dark scenes, the a.g.c. voltage diminishes and there is a tendency for the background to show up because of the increased gain. This is particularly annoying in weak-signal areas where the background may already be rendered brighter than intended by the presence of noise on the signal. The third objection is that the available voltage from this system is inadequate to allow a delayed system of control on the r.f. stage.

* Fesguson Radio Corporation.



Fig. 4. Standard sync waveform showing (a) a line sync pulse and (b) part of frame sync pulses.

Fig 2. shows the basis of frame-pulse gated a.g.c. This method gates the valve producing the control voltage so that the output depends on the amplitude of the sync pulses during the frame fly-back, and since picture components do not occur during this period, the output is a true measure of the signal level. The contrast control may operate to determine the amount of pulse amplified by the a.g.c. valve. Because the gating frequency is so low, a long time-constant is required in the smoothing circuit. The frame hold control must be correctly set before the contrast to ensure that the gating waveform occurs at the right instants

For the line gated system shown in Fig. 3 the gating pulse is derived from the line time-base, and is delayed so that it operates the a.g.c. system during the "back porch" period of the line sync pulse. Since the back porch does not exist at the gating times during the

frame pulses, i.e., approximately 12µsec after the negative-going edges in Fig. 4(b), there is a loss of a.g.c. during these periods, making it necessary to use a fairly long time-constant in the smoothing circuit. There is interaction between the line hold and contrast controls, because the system relies on the line time-base timing to obtain gating pulses at the correct Usually these instants. pulses are taken from the output circuit of the timewhich does not base, operate for some time after the rest of the set has been switched on because of heating delay in the effi-ciency diode. Thus a protection circuit is required for this period to prevent overloading of the video circuits.

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The "Sync-Gated" System .- The new form of a.g.c. system is basically a gated system, controlling the amplitude of the signal during the "back porch" of the line sync pulse, but the gating pulse is derived from the sync pulse itself rather than from the timebase, thus overcoming the difficulties experienced with some of the other systems as a result of the interdependence of "hold" and contrast controls. Reference to Fig. 4, showing the standard 405-line television synchronizing waveform, will explain the principle of the system. If this waveform is examined it can be seen that there is always a period of black level immediately following the positive-going edge of a sync pulse, whether it be a line or a frame pulse. Hence, a short-duration pulse derived from this edge and delayed slightly so that it occurs a microsecond or two after the edge, would form a suitable gating pulse for an a.g.c. system.

This is the basis of the new arrangement, the required gating pulse being obtained simply by differentiating the sync pulse and making use of the positive half of the waveform so obtained. This pulse is then added to the video waveform, the total amplitude being equal to the sum of the original pulse and the "back porch" level of the video signal (see Fig. 5). This total amplitude is made always to exceed the peak video amplitude and is passed through an amplifier, the pulses appearing at the anode of the amplifier being peak-rectified and used as a.g.c. voltage.

A very important feature is that the pulses are still present during the frame sync intervals—in fact they occur at twice the line frequency during this period and so the considerable fall in a.g.c. voltage which occurs with line time-base gated systems during the frame sync interval is replaced by a small increase in voltage which can, however, be kept to a minimum by operating the rectifier under, as nearly as possible, peak rectifying conditions. In actual practice it has been found that, for a given percentage of 50-c/s ripple on the a.g.c. line, a considerably shorter a.g.c. time-constant can be used with this system than with the line time-base gated system, thus improv-



Fig. 5. Circuit of the new sync-pulse gated system.

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ing control of rapid fading such as aircraft flutter.

It is, of course, essential that the sync separator in the receiver is capable of maintaining its output sync pulse constant in amplitude and speed over an adequate range of contrast settings, otherwise a positive feedback loop will exist via the a.g.c. line, and lowfrequency fluctuations will occur in the gain of the controlled stages of the receiver. In commercial receivers the sync separator usually limits inadequately below about 10 volts peak-to-peak video signal, with the result that these fluctuations may occur if the contrast control of the receiver is adjusted to give a video voltage of this order at the cathode (or grid) of the cathode-ray tube. This is not considered to be a very serious objection, since such an amplitude of video signal represents a picture of extremely low contrast on the average cathode-ray tube.

A practical circuit is shown in Fig. 5. The video waveform including sync pulses is taken from the video stage cathode and a differentiated pulse from the sync separator (V1) anode, formed by the short timeconstant C_1R_1 , is added to it at the grid of the a.g.c. amplifier V2. Since the cathode of the video stage is a low-impedance point, the amount of pulse fed back into the video amplifier is very small. The delay in the sync separator is sufficient to ensure that the positive part of the pulse lies over the back porch.



Fig. 6. A.G.C. curves for Ferguson Model 203T receiver with various settings of the contrast control. The video waveform amplitude is peak-to-peak (including sync) measured at the cathode of the video amplifier stage. It should be multiplied by 23 for the amplitude at the cathode-ray tube.

Fig. 7. A.G.C. curves for the 203T receiver showing the distribution of control between r.f. and i.f. stages.



The valve V1 is operated under very short grid-base conditions so that the pulse produced is constant in amplitude over a wide range of contrast, and this amplitude is sufficient to extend well beyond the peakwhite level at the grid of V2. This valve is biased by the contrast control R_3 and resistor R_2 so that it conducts only on the tips of the pulses. Hence the output at the anode of V2 consists of pulses whose amplitude for any given contrast setting depends only on the back porch level and therefore the signal level.

These output pulses are coupled to the diode V3 circuit by C_3 and R_5 . The diode load is split into R_6 and R_7 , the tapping supplying bias to the vision i.f. stage. A delay network comprising R_8 , R_9 and the diode V4 supplies bias to the r.f. stage. The capacitors C_4 and C_5 are small, so that a short time-constant exists on the control line of the vision i.f. stage. Since there is a small change in the a.g.c. voltage during the frame pulses, a longer time-constant is used for any stages common to vision and sound. Thus extra smoothing, R_{10} and C_6 , is added for the r.f. stage.

The relative values of the resistors forming the delay system may be found in the following manner. If e_1 is the required i.f. stage bias at which control on the r.f. stage should commence and, under maximum signal conditions the i.f. stage bias should be e_2 when

the r.f. stage bias is e_3 in order to avoid overload of the early receiver stages, then we may write:—

 $Ae_1 - Be_0 = 0$ $Ae_2 - Be_0 = e_3$

where

and

$$A = \frac{R_6 + R_7}{R_7}$$
$$B = \frac{R_8}{R_8 + R_9}$$

 $e_0 = h.t.$ voltage These equations yield:

and

$$\mathbf{B} = \frac{1}{e_0} \cdot \frac{e_1 e_3}{e_2 - e_1}$$

 $A = \frac{e_3}{e_2 - e_1}$

Curves showing the effectiveness of the system as applied to a complete receiver are given in Fig. 6 and it can be seen that for an average setting of the contrast control (the curve marked $V_k = 14.5$) a change of 60dB in input signal is reduced to only 2.5dB at the cathode-ray tube. At a higher contrast setting ($V_k = 15.2$) the control is not quite so good but even so a change of 60dB in input causes only 8dB change in video voltage.

Fig. 7 shows the way in which the control over the total receiver gain is divided between the r.f. and i.f. stages. This clearly demonstrates how the r.f. gain is maintained at a high level for weak signals in order to achieve a good signal-to-noise ratio, while with strong signals the majority of the gain reduction occurs in the r.f. stage in order to avoid overloading the mixer stage or first (uncontrolled) i.f. amplifier.

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Information Theory

Widening Fields of Application Discussed at London Conference

HE third symposium on information theory to be held in London* was housed this year at the Royal Institution. The change of address from the Institution of Electrical Engineers, where the last one was held, certainly reflected the different character of this year's proceedings, which seemed to have kicked away the ladder of mere technology by which information theory rose to fame. Whereas the I.E.E. symposium was mainly concerned with practical applications of Hartley's and Shannon's work and had a direct appeal to communications engineers, the Royal Institution offered such a miscellaneous collection of scientific subjects that any engineers present could have been forgiven for asking themselves what they were doing there at all. Welsh spelling rubbed shoulders with nerve mechanisms, randomizing machines with the cochlear nucleus and games strategy with optical transmission, not to mention statistics, translating machines, codes, linguistics, clerical operations and extra-sensory perception.

There were, however, a few topics that could be said to represent our own sphere, and one of particular interest was concerned with coding information to increase the efficiency of communications channels. This term "coding" is often used loosely to describe a process by which the signal is converted from one physical variable into another (for example, from amplitude changes to frequency changes, as in f.m.) without altering its essential structure or pattern. What "coding" should really mean is a process which takes into account the statistical nature of the signal, so that less time or signalling power is given to representations which occur frequently, and therefore convey little information, and more to those which occur infrequently and give a good deal of information. A familiar example is the Morse code, where the letters which have the greatest probability of occurring in English text, like E and T, are given the shortest symbols.

A good deal of research is being done to apply this kind of coding principle, not only to "telegraphic" signals but to continuous ones like speech and television waveforms, in order to reduce the bandwidth or signalling power required in communications channels. It becomes particularly valuable when the "conditional probabilities" are taken into account. For example, in English text when the letter Q occurs it is almost certain to be followed by U, so that when the U actually comes up in these particular circumstances it conveys very little information and should be given less time or signalling power than it normally gets. Again, this principle can be applied to signals of a continuous nature.

A more recent offshoot of this work has been on special codes for detecting and correcting errors, and these are used mostly for "telegraphic" signals such as the trains of pulses in digital computers or telephone dialling systems and, of course, in telegraphy itself. The term "errors" here refers to changes in the signals caused by noise, intermittent faults or any other kind of interference with the transmission chan-

* Organized by Dr. Colin Cherry, of Imperial College.

nel. In a digital computer, for example, a single noise pulse added to a train of binary digit pulses representing a number could make a difference of thousands or millions in the result of a calculation. The general principle of detecting and correcting such errors is to add some extra symbols to the "message" to act as a check on the accuracy of what is received, rather like the tally of words sent in a telegram. This entails some redundancy in the code and its efficiency is thereby reduced, so the aim of the "information theorists" is to design effective codes with minimum redundancy. The methods used for detecting the errors are complex, but in the binary system of signalling (with which the codes are mostly concerned) the process of correction is simplicity itself. Since there are only two symbols used, 0 and 1, once it has been determined that an error exists at a particular position in a train of digit pulses it is only necessary to reverse the digit in that position from 0 to 1 or from 1 to 0 as the case may be.

Speech Recognizing Machine

Papers concerned with various aspects of binary coding at the symposium were "The Synthesis of Linear Sequential Coding Networks" by D. A. Huffman, "A General Class of Codes and their Physi-cal Realization" by A. E. Laemmel and "Coding for Noisy Channels" by P. Elias. The notion of error correction and that of "conditional probabilities" mentioned above also came into an interesting description by D. B. Fry and P. Denes of their experimental work in building an automatic speech recognizer-a machine that will produce typewritten versions of words spoken into it. The main problem here is that a machine cannot recognize words simply by their acoustical properties, which vary considerably with pronunciation, intonation and quality of voice. This is also true of *homo sapiens*, of course. It is necessary for the machine, like the human being, to have a store of linguistic information to act as a check on the acoustical frequency patterns and modify them when they seem to have an improbable structure. In the recognizer described by Fry and Denes this is done on an elementary level by using the "conditional probabilities" with which certain sounds occur after other sounds in a given language.

First of all, the purely acoustical patterns are "recognized" by a series of filters, which divide up the frequency spectrum into adjacent bands. Most phonemes give rise to characteristic spectral patterns which can generally be identified by coincident peaks of energy in particular pairs of filters. The outputs of the various pairs of filters are multiplied together and this gives a series of voltage products which vary in relative magnitude as changes in the applied speech wave move the energy maxima from filter to filter. An electronic circuit then detects the greatest of these products and indicates the occurrence of the corresponding phoneme.

The linguistic information on "conditional probabilities" is stored in the form of slider adjustments on

sets of potentiometers. Each set is associated with a particular phoneme and the various sliders give voltage outputs proportional to the probabilities of the other phonemes which may follow it. What happens in operation is that when a particular phoneme is recognized by the filter-pair system, say i, this information is stored in a corresponding memory circuit until the next phoneme occurs and then used to switch the voltage indicating the occurrence of this next phoneme to the i set of potentiometers. The "probability voltages" from the sliders are then sent to the filter-pair system, and here each phoneme acoustical output voltage is multiplied by its corresponding probability voltage. Thus the acoustical recognition of the phoneme following the i is biased or weighted according to the probability of its occurring in the language. So if the filter-pair system has a tendency to produce the largest voltage for a phoneme which is similar to the one spoken but not actually correct, the weighting by the probability voltages will most likely swing the balance and cause the correct acoustical output to have the largest voltage. This is then selected by the electronic circuit mentioned above and used to operate an electric typewriter.

Some interesting and amusing typewritten results were shown by Fry and Denes which clearly demonstrated the greater accuracy of recognition obtained with the aid of the store of linguistic information.

Applications of Computers

In one sense mechanical speech recognition can be regarded as a coding problem since the actual structure of the information is altered. A rather different type of coding, though again for transformation of language information, was described in a paper by A. D. Booth and J. P. Cleave-the "programming" of a digital computer to transcribe ordinary alphabetic text into Braille. The technique was similar to that outlined in the article on language translation in a recent issue* except that with Braille the transcription was in terms of individual letters and simple groups of letters (or simple words) called "contractions." There is not, however, a simple one-to-one correspondence between printed characters and the Braille signs, for the particular signs used depend upon the context. Transcription into contracted Braille, in fact, is almost as complex as translation into a foreign language, except that the words in real language are replaced by "letters" in Braille and the word groups or idioms by Braille letter groups.

Computing machines are also being used in another field of information theory to investigate the transmission of information in the animal (or human) nervous system, which can be regarded as a kind of communications network. The general idea is to organize the machines to imitate the external patterns of behaviour produced by the nervous system in the hope of throwing some light on the actual physiological mechanism of the system, which is very difficult to investigate by One paper, by O. G. Selfridge, other means. described the adaptation of a digital computer to give recognition of visual patterns. Here, there was no attempt to make the machine imitate the nervous structure directly-only its functions. Another paper, however, by W. K. Taylor, described a network more in the nature of an analogue computer which was built up from model neurons having electrical proper-

* "Language Translation by Electronics," by J. P. Cleave and B. Zacharov, Wireless World, September, 1955. ties similar to those of living nerve cells and sensory receptors. Each model cell contained electronic circuitry using seven valves. Experiments were described in which the correct "physiological" response to a pattern of stimulation could be "learnt by association" by the network.

In particular the sense of hearing offers a challenge to information theory because of the complexity of the path from air to brain and because of its incredible sensitivity. It has been calculated that at the threshold of hearing the displacement of the eardrum is less than the diameter of a hydrogen molecule. If the total energy available is divided by the observed number of sensory cells in the inner ear it turns out that the signal is well down into the thermal noise level. This results in a constant random "firing" of nerve threads which is to some extent ordered by interconnection, feedback and autocorrelation in the presence of even the weakest sound stimulus.

A detailed investigation of the function of the cochlear nucleus, reported by J. T. Allanson and I. C. Whitfield, confirms this view and shows that the pattern of response to a pure tone is modified from a Gaussian or humped-back distribution of activity in the ingoing fibres, to a steep-sided, flat-topped bandpass response in the outgoing fibres. This sharpening of the region of neural activity is thought to account for the observed degree of discrimination of pitch, and it was suggested that the width of the "passband" might prove to be related to the perception of intensity. Weight is lent to this hypothesis by the known dependence of subjective pitch on intensity.

Frequency analysis lies at the root of all theories of hearing and a comprehensive survey of the present state of our knowledge was made by J. C. R. Licklider who pointed out that the "place" theories of Helmholtz and Bekesy, which identify specific regions of excitation in the cochlea with different frequencies, were insufficient to explain the subjective recognition of a low-pitched fundamental residue when the ear is stimulated by high harmonics only of that fundamental. Schouten's work on this subject had made it clear that the low pitch remained after any combination tone due to non-linearity in the ear had been balanced out by a variable-phase search tone.

An alternative demonstration was given by Prof. Licklider in which short bursts of a sine-wave fundamental were interspersed with tones consisting of the higher harmonics only. A random low-frequency masking noise was then introduced, which completely suppressed the sine wave but left the sensation of low pitch from the harmonics untouched.

These results could not be explained by any existing simple "place" theory. There was the possibility that a transformation from time to place might occur in the cochlear nucleus or at high levels in the path from ear to brain, and the main part of his paper was devoted to a detailed description of a "triplex" mechanism which might fit the facts so far observed.

It seems, then, that the principal role of information theory in all these different fields has been to throw a new light on old problems. Channels of communication have been found to exist in situations which were never considered in this way before. At the same time one cannot escape the conclusion that many of the researches described at the symposium might have gone on quite successfully without information theory! However, the world of communications must take it as a compliment from the rest of science that such attention is being paid to the subject.

Measurement of "Wow" and "Flutter"

Basic Circuitry of a Test Instrument

DEFORE the subject of measurement can be considered it is necessary to form a clear picture of what exactly is meant by "wow" and "flutter," as misunderstandings often arise where terminology is concerned.

Wow or flutter is a periodic or non-periodic frequency deviation (occurring mostly when reproducing from magnetic tape or disc records) expressed as a percentage of the mean frequency. For example, if a recorded signal of 3,000 c/s varied on reproduction between 2,970 and 3,030 c/s, wow or flutter in this case would be:

$$\frac{3,030-3,000}{3,000} \cdot 100 = \frac{3,000-2,970}{3,000} \cdot 100 = \pm 1\%$$

We are here dealing with peak-to-peak values, as we should in the case of slow deviations, but for

higher rates the mean value of frequency deviation is usually considered. Consequently, if flutter were given as, say, 1% without saying that this referred to peak-to-peak deviation, it would be safer to take the maximum frequency range as being appr

frequency range as being approximately $\pm 1.5\%$, assuming that the waveform is nearly sinusoidal.

A second term (entirely different in meaning from percentage wow or flutter) is wow/flutter *frequency*. As we mentioned previously, change of frequency may often occur in a periodic way. This period of frequency variation can be from a few cycles/sec (or even less) up to a few hundred cycles/sec, and can be divided into two ranges, i.e.: "wow"—corresponding to frequencies within the limits 0-20 c/s and "flutter"—covering all frequencies between 20 and 200 c/s. Wow itself could be divided again into a "very slow" wow, 0.2 c/s* (not so audible), and "normal" wow within the limits 2-20 c/s. It is significant that "very slow" and "normal" wow and flutter are different not only audibly but also from the measuring equipment point of view; a "very slow" wow has to be measured by different circuitry.

Choice of Carrier (Test) Frequency. Frequency

*British Standard 1988:1953 recommends the use of the term "drift" for deviation of frequency "below about 0.1 cycles per sec." [Editor.] variation when occurring with the same percentage deviation over a wide range of carrier frequencies has different audible effects depending upon carrier frequency, rate of change (wow/flutter frequency) and individual hearing characteristics. It can be assumed, however, that a carrier frequency lying between 2,000-5,000 c/s is most vulnerable to wow and flutter and consequently 3,000 c/s has been chosen internationally as the test frequency.

Purpose of Wow/Flutter Measurements. There are two main reasons for this sort of measurement. First, to check the percentage of frequency variation to ensure that it does not exceed the allowed limit; and, if this is the case, to make necessary adjustments and repairs in the apparatus under test. Secondly, to establish an objective figure of merit which can be used also for equipment with frequency deviations

which are difficult of detection by the average human ear. For instance it can be generally assumed that equipment with w/f(wow or flutter) higher than 0.5% represents poor performance and below 0.05% is associated

with best-quality reproducers. As flutter meters are most used for fault finding, it often happens that not only percentage of w/f has to be checked, but also w/f frequency estimated—the chief reason being to find the cause of the trouble. This certainly complicates matters, particularly as the flutter waveform is generally of a complex nature. We shall return to this subject later in the article. **Principle of Flutter Meter.** Fig. 1 is a block diagram of a w/f analyser. Flutter phenomena are

similar to frequency modulation in r.f. circuits. Consequently the flutter meter has similar stages to those of a f.m. receiver, the chief difference being, of course, an audio-frequency carrier (3,000 c/s) and very low deviation. This latter factor has an important influence on the design of the limiter and output stages.

Let us consider stage-by-stage how a flutter meter works. The signal of nominal frequency 3,000 c/s, from the output of, say, a tape recorder, is fed through an amplifier to a limiter. The purpose of the limiter is to reduce the minimum amplitude





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O. E. DZIERZYNSKI

referred to O. E

modulation, and to provide a reasonably constant input to the discriminator.

The limiter is followed by a band-pass filter which restores the sinusoidal waveform which was lost in the limiter stage. The level here is of the order of 100-500 mV.

The discriminator is designed as an amplifier with a linear (or close to it) characteristic of gain with frequency within the limits 2,700-3,300 c/s. An ordinary parallel-tuned L-C circuit in the anode circuit of the discriminator is good enough for this purpose and this circuit should be tuned to approximately 3,400 c/s (peak) (see Fig. 4). Under these conditions, the discriminator amplifier would have only about half of the available gain from the carrier of 3,000 c/s (working on the slope of resonant curve).

If a frequency-modulated 3,000 c/s signal is applied to the grid of V3 the output from the anode will be an amplitude-modulated carrier (incidentally still partly frequency modulated). This produces an amplitudemodulated carrier which should, if necessary, be amplified to at least 10 V before detection.

Diode detection is usually employed and the output from it, after removal of the d.c. component,

Vib

5kΩ

VIa

P 500kΩ

ıkΩŞ

0.05 µ.F

╢

0.1µF

╢

ECC 81

25 LL F

+ H.T.

600Ω

Fig. 2. The input

stage comprises a

pre-amplifier, cath-

ode follower and

3 kΩ ₹ ||

0.04 µF

0·02μF 0·02μF

0.04 1

lo-3H

3kΩ

high-pass filter.

500 k Ω

10k Ω

.....

contains the w/f frequency plus the remains of the carrier and carrier-harmonic frequencies.

As the flutter component is often a matter of only a few millivolts, a low-frequency amplifier has to be introduced between detector and a.c. output meter. The gain of this amplifier is of the order of 35-40 dB to ensure that the a.c. output meter (about 10 V f.s.d.) is working on the linear part of its scale.

A low-pass filter (cut-off frequency about 300 c/s) is often inserted between detector and output meter, to reduce carrier-frequency components.

As we shall see later, provision can be made to measure very low frequency wow (or frequency drift) by inserting a d.c. microammeter in series with the diode load resistor R_L (Fig. 4 (a)). Variations of frequency will be followed by the instrument pointer, provided that they are slow compared with its natural period, i.e., below about 2 c/s.

Circuit Details

Input Stage. Fig. 2 gives more details of the input circuit. The reproduced test signal of 3,000 c/s is passed through the pre-amplifier V1A, cathode follower V1B and high-pass filter. All lower frequencies (including hum) present in the output of the tape recorder are removed, as they can otherwise affect the final reading of the output meter. The purpose of the cathode follower is to match the low input impedance of the filter (3,000 ohms) to the high output impedance of the pre-amplifier. The choice of a low filter impedance has the advantage of making the design of the filter much easier, also

TO LIMITER

INPUT

of reducing hum pick-up at the input of the limiter. Limiter. Two types of

Limiter. Two types of limiter circuit were considered during the development of the meter described in this article. The first employed a ECC81 working as an overloaded cascade amplifier with additional clipping action by diodes. A square valve was ob-



+ H.T



Fig. 3. (a) Circuit of limiter and band-pass filter stage; (b) input/output characteristic; (c) waveform at A when V_A exceeds the bias V_b .



Fig. 4. (a) Discriminator and detector circuit; (b) Microammeter connections for measurement of drift; (c) Effect of R_{α} on slope of discriminator curve.

tained of peak value 75 V, approximately constant for all values of input voltages within the limits 80 mV-10 V. Fig. 3 (a) shows the second system which was finally adopted. Two negatively biased diodes are fed symmetrically with the a.c. signal through resistor R_1 . At A the waveform is sinusoidal until the peak value of the signal exceeds the bias when the output takes the form shown in Fig. 3 (c). The amplitude remains substantially the same for inputs varying within similar limits to those quoted for the first system.

Band-pass (Sine-wave Restoring) Filter. Several systems have been considered and for simplicity and efficiency a single tuned circuit was chosen fed by resistor r. Damping of this tuned circuit had to be

chosen very carefully, as sidebands within the limits $3,000 \pm 200 \text{ c/s}$ had to be passed without appreciable attenuation. At the same time attenuation of the second and third harmonics (6,000 and 9,000 c/s) should be 20 dB at least. Fig. 3 (b) represents the limiter and filter characteristic, i.e., how the output of this stage depends upon input. There is a broad maximum for the input values 20-30 mV and it is clear that the optimum input for minimum amplitude modulation is about 25 mV.

Discriminator and Detector. Any r.f. pentode with sufficient gain could be employed for the discriminator, which is coupled to the detector stage. In the circuit of Fig. 4(a) the discriminator is the heptode section of an ECH81, which is of the variable mu-type



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Fig. 5. (a) Circuit of output stage and (b) typical frequency response.



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To stabilize the gain the triode section is connected to the h.t. to increase the cathode current so that a lower value of cathode bias resistor can be used.

It is important that the slope of the resonance curve should be sufficient to produce about 1 volt of flutter frequency across the detector load R_L for an input to the grid of V3 of about 500 mV with a frequency deviation of 0.5% (of 3,000 c/s). By shunting the resonant circuit (R_x) it is possible to reduce the discriminator efficiency very considerably. In this way it is possible to obtain a change-over to different ranges (see Fig. 4(c)). In this particular case resistors of 45 k Ω and 12 k Ω would cause a reduction of slope of 3 and 10 times respectively, while 1 volt a.c. of detector output could still be obtained from frequency variations of 1.5% and 5% respectively (with a signal of 500 mV fed to the discriminator grid). It will be seen from Fig. 4(c) that the value of rectified carrier current *i*₀) will be different in each range.

Very slow wow is measured with a microammeter (Fig. 4(b)) inserted in the detector load circuit. With a load resistor of $0.25 M\Omega$ current i_{L} is approximately $180 \,\mu$ A, and a frequency change of $\pm 0.5 \,\%$ would give a current change of approximately $\pm 5 \,\mu$ A. Employing a meter with f.s.d. $100 \,\mu$ A, it is possible to compensate $130 \,\mu$ A from an external source to obtain neutral indication exactly in the middle of the scale. If the scale is calibrated in cycles, peak-to-peak measurement of the slow swing can be made. This method has the advantage that large slow variation of frequency can be measured, without range-changing facilities, as, even with wow up to $\pm 5\%$, movement of the needle would be within the limits of the scale.

Output stage. As explained previously, a lowfrequency amplifier and filter for carrier components must be included in the last stage of the flutter meter. Fig. 5 shows how this problem was solved in a simple and satisfactory way. The first half of the ECC83 is a R-C amplifier while the second triode serves as an output cathode follower feeding the output meter through a special network. A cathode rejector circuit tuned to 3,000 c/s was employed to reduce any remaining carrier signal (it was found that second and third harmonics were negligible at this point in the circuit).

The frequency response of the amplifier was given a strong lift towards 200 c/s (pre-set potentiometers P_1 , P_2 ; condensers C_1 , C_2) to compensate considerable loss of "higher" frequencies (in a band 5-200 c/s) through attenuation of sidebands in the discriminator circuit and smoothing capacitances in the detector circuit. The resulting frequency response (Fig. 5(b)) is more or less uniform: there is some loss at frequencies between 100-200 c/s and 5-10 c/s, but experience shows that the most important flutter frequencies are within the limits 10-100 c/s.

Oscillator and Ancillary Circuits. In Fig. 6 the EF86 is a stabilized 3,000-c/s oscillator with one half of the ECC81 as a buffer valve, while the second half works as a relay valve (see later). There are also some facilities for setting the exact level and for analysing flutter frequency contents. The oscillator circuit is a conventional Colpitts electron-coupled oscillator with automatic limitation of amplitude (see Wireless World, March 1954, p.110). The output is fed to a capacity attenuator which is used to maintain good waveform on lower ranges. Output voltages are in decade steps ranging from 1 mV to 1 V.

The second section of the ECC81 (V6b) is used as a relay valve. It is found that the oscillator, unless well screened, introduces a beat note with the played-back signal which upsets the accuracy of measurement. Consequently, it was arranged that while the recorded signal is played back, the oscillator is automatically switched off. This was achieved by applying a positive bias to the grid of the relay valve, this bias being derived from rectification of the 3,000 c/s played-back (fed from anode circuit V2b—see Fig. 3(a)). Only when signal is being passed through the limiter of sufficient level to give an anode current in



Fig. 6. Test oscillator (3,000 c/s) and other auxiliary equipment.



the relay of about 3 mA, does the armature of the relay close contacts c-c, shorting to earth the grid of the oscillator. Another pair of contacts b-b lights up a pilot bulb, indicating that the test signal is of approximately the right level. Accurate adjustment of level (to minimize a.m.) can be made by switching the output meter to the anode circuit of the limiter (point B), where the signal should be 3 volts, and setting the input potentiometer P (Fig. 2). This adjustment of level is recommended when making an ordinary test of flutter, and essential when frequency drift is checked. We shall return to this subject later.

Flutter Analysis. Flutter wave form is usually complex and irregular. It often happens that one frequency is predominant and it is desirable to estimate this frequency. Fig. 5 shows how this frequency may be found by using an external calibrated lowfrequency generator. On the reduced flutter output (say 1/5 of f.s.d.) is superimposed about 20 V a.c. from the external generator, resulting in a deflection of the instrument to some point near the middle of the scale. Switching over to the external oscillator is carried out by switch S which simultaneously reduces the flutter level by introducing resistor R.

If now the frequency of the external generator is slowly changed from 10 to 200 c/s, the deflection will not change until one of the flutter frequencies approximates the external frequency. The needle of the instrument then indicates a slow beat with the search frequency. This is well known as the simplest method of analysing a complex wave, but in our case, a search often has to be repeated several times, as flutter usually has varying amplitude and it is easy to miss the beat frequency.

Frequency Drift. In tape recorders very slow frequency drift is sometimes experienced. This can be checked and measured by comparing the rectified

3,000-c/s carrier from the test oscillator and the played-back recorded note with the auxiliary meter Levels of both signals have to be of Fig. 4(b). identical and the following procedure in this test has to be observed (see Fig. 6). While testing, the level of the played-back note has to be adjusted by poten-tiometer P (Fig. 2) with button "a" depressed (output meter set to f.s.d.). Then, depressing buttons "b" and "a" simultaneously, the output from the oscillator is fed to the input socket of the flutter meter and f.s.d. adjusted by fine control of the attenuator. In this way both levels are set identically. Then with push buttons "a" and "b" off and recorded tone 3,000 c/s still going through, the reading should be noted on the microammeter (Fig. 4(b)). By depress-ing only button "b" the carrier from the oscillator is in turn measured. If both frequencies are the same (no frequency drift), both readings should be the same; the usual procedure is to make a note of the second reading (oscillator), switch back to the signal from the recorder and observe subsequent drift. The microammeter scale can be calibrated to read directly in cycles per second of drift.

Calibration of Flutter Meter. Although initial alignment can be made using a convention a.f. signal generator, calibration of gain requires the use of a flutter generator. For our purpose the flutter generator should produce a frequency-modulated 3,000 c/s carrier, with frequency deviation continuously controlled up 10%, at repetition rates within the limits 2-200 c/s.

As a general guide Fig. 7 gives a block circuit of the calibrating arrangements. This is basically a beat-frequency oscillator with a fixed frequency 42 kc/s supplied from oscillator A (small amplitude) and variable frequency from oscillator B (39 kc/s large amplitude). After mixing and detection the

resulting beat note (3,000 c/s, frequency modulated) is passed through band-pass filter C. The chief purpose of this filter is to remove from the output traces of superimposed modulating frequency and to suppress higher frequencies derived from oscillators A and B. Frequency deviation is achieved by a reactance valve circuit in which deviation limits are controlled by the l.f. oscillator D (voltage measured by V.V.). In this case the reactance valve, with initial bias 3.5 V, would produce a change of frequency in oscillator B within the limits ± 150 c/s with a modulating voltage of 640 mV r.m.s. Consequently the output would be $3,000 \pm 150$ c/s—which represents 5% flutter (peak-to-peak). For deviations of 1.5 and 0.5%, 195 mV and 67 mV r.m.s. respectively were required from the modulator.

Commercial Literature

Junction Transistor Circuits for home construction using Mullard OC70 and OC71 types. Brochure giving basic circuit configurations, operating instructions, explanation of data, characteristic curves and various practical circuits from Mul-lard, Century House, Shaftesbury Avenue, London, W.C.2.

Wire-wound Potentiometers and rheostats, including ganged and concentric types. Specifications and mechanical details in an illustrated catalogue from the British Electric Resistance Company, Queensway, Enfield, Middlesex.

Amplifier Printed Circuit for the Osram "912" audio amplifier. Instructions for assembling the components on the printed circuit and details of a suitable classis given in an illustrated technical bulletin from The Telegraph Condenser Company, North Acton, London, W.3.

Light-weight Flexible Cords for hearing-aids, headphones, microphones, etc. Table giving details of cores, coverings, resistances, current ratings and other information on the range produced by Amplivox (Industrial Products Division), 2, Ben-tinck Street, London, W.I.

U.H.F. Communications Receiver covering 150-500 Mc/s (a.m. and f.m.) in six positions of a miniature tuning turret, with sensitivity of $10 \ \mu$ V for 15-dB signal/noise ratio. Can be powered from batteries as well as mains, and facilities are provided for use as a test instrument. Illustrated leaflet on the Eddystone Model 770U from Stratton & Co., Alvechurch Road, Birmingham, 31.

Special-purpose American Valves.—A list giving type num-bers and descriptions of the types available from Industro, 649, Broadway, New York 12, N.Y., U.S.A.

Aluminium Soldering Tool using a vibrating steel-wire brush in the centre of the soldering bit. Descriptive leaflet from the Belark Tool and Stamping Company, 130, Mount Street, London, W.1.

High-quality Audio Amplifier with 12 watts output, fre-quency response of 20 c/s-25 kc/s within 0.2 dB and total harmonic distortion of less than 0.1% at 10 W and 700 c/s. Disc record compensation and variable 1.s. damping factor. Also a pre-amplifier control unit. Leaflet from R.C.A. Photo-phone, Lincoln Way, Windmill Road, Sunbury-on-Thames, Middlered Middlesex.

Gramophone Turntable Unit of "transcription" quality for three-speed operation, claimed to have no wow, futter, rumble or hum induction. Descriptive leaflet from Woollett Sound and Wireless Equipment, Wells Park Road, London, S.E.26.

Distributor's List of products stocked and manufacturers. Available to trade, research and industrial organizations from Holiday and Hemmerdinger, 74-78, Hardman Street, Deans-gate, Manchester, 3.

Plastic Extrusions for insulation purposes, mostly in P.V.C. Leaflet outlining manufacturing facilities available, from Creators, Plansel Works, Sheerwater, Woking, Surrey.

R.F. Induction Heaters, bench type, with continuously variable heat output over a 20:1 range. Two models available: 1 kW and 3 kW. Specification on a leaflet from Radio Heaters, Eastheath Avenue, Wokingham, Berks.

Transcription Gramophone Unit with speed continuously variable from above 78 r.p.m. to below 16 r.p.m. Also ver-tical-edge driving pulley claimed to eliminate rumble. Leaflet from The Goldring Manufacturing Company, 49-51A, de Beauvoir Road, London, N.1.

Multi-pole Changeover Relays; sealed types suitable for air-craft installations, with novel wiping action of contact sur-faces and higher voltage contact blocks. New developments in Pullin relays described in a leaflet from Measuring Instru-ments, Electrin Works, Winchester Street, Acton, London, W.3.

Secondary Cell, 2-volt, for use with vibrator power supplies of portable equipment. Notable improvements in plastic con-tainer and combined gas vent and acid trap. Information from Chloride Batteries, Exide Works, Clifton Junction, Swinton, Manchester.

SHORT-WAVE CONDITIONS Predictions for November



THE full-line curves given here indicate the highest frequencies likely to be usable at any time of the day or night for reliable communications over four long-distance paths from this country during November. Broken-line curves give the highest frequencies that

will sustain a partial service throughout the same period.

FREQUENCY BELOW WHICH COMMUNICATION SHOULD BE POSSIBLE FOR 25% OF THE TOTAL TIME

- --- PREDICTED AVERAGE MAXIMUM USABLE FREQUENCY

FREQUENCY BELOW WHICH COMMUNICATION SHOULD BE POSSIBLE ON ALL UNDISTURBED DAYS

Polarization of Waves

By "CATHODE RAY"

What it Means and How it Can be Used

HAVE a feeling that wave polarization is one of those things that the non-specialist tends to regard as too difficult to be worth trying to understand. When he comes across the news—as he may easily do nowadays—that radar is being rendered more effective by the use of circularly or elliptically polarized waves, does he have a clear picture of what this means? Or is there a certain amount of mental haze?

Many explanations of the polarization of radio waves conveniently assume that polarization of light is already understood. But so far as we are concerned it will probably be easier to understand the polarization of light in terms of polarization of radio waves, rather than vice versa. So I shall assume only a reasonable basis of radio, and then anybody who does happen to know about optical polarization will be able to amuse himself by looking out for the analogies.

First of all, then, let us refresh our memories about the main features of radio waves.



Fig. 1. Time-honoured procedure for generating transverse rope waves, in this case vertically polarized.

Unlike sound waves, they are transverse waves. That is to say, they consist of vibrations or alternations across the direction in which they are travelling. So they can be pictured quite aptly by the juvenile pastime of waving a long outstretched rope from side to side or up and down (Fig. 1). This process causes waves to travel along the rope—in spite of the fact that the movement which causes them is at rightangles to that direction. These rope waves could be called positional waves, because the thing that alternates is the position of each bit of the rope. If the rope's end is waved up and down, the position of that and every other bit of the rope alternates vertically and the resulting waves could be described as vertically polarized. If it is waved from side to side, they are horizontally polarized. The meaning of polarization is as easy as that!

The corresponding thing about radio waves is a little less easy, because it can't actually be seen. More-

over, in radio waves two things are alternating at once -electric field and magnetic field-both of them across the path of the wave, but at right angles to one another (Fig. 2). So although even the most perverse scientist could hardly refer to the polarization of his son's up-and-down rope wave as other than vertical, there is room for difference of opinion on what constitutes a vertically polarized radio wave. And where there is room for difference of opinion in scientific terminology, believe me, full advantage is almost invariably taken of it, to the confusion of all concerned. This case is no exception. In defining the direction of polarization there is a natural choice between two alternatives: making it the same as the direction of the electric field or of the magnetic field. Which has been chosen? Both! Electric for radio waves and magnetic for light waves, though radio and light are identically the same things except for frequency!

We radio people can argue, in favour of our choice, that it is so convenient to be able to say that a vertical dipole radiates and receives vertically polarized waves. The optical types, on the other hand, can retort that they were first in the field (even before anyone realized that there was a field!) and how could Mr. E. T. Malus (who invented "polarization") have foreseen in 1808 what TV fans would find convenient in 1955, and that it would have anything to do with the peculiar optical effects he was noticing in crystals of Iceland spar?

Anyway, from now on let us simplify matters by taking the "field" of radio waves to mean their electric field. Although we know the magnetic field is there too, for the present purpose we can leave it out of account.

Unlike generators of light—commonly called lights —generators of radio waves usually start the waves off with a definite direction of polarization. Although we have mentioned only two varieties—vertical and horizontal—because these are the only two in popular use, the angle of polarization can, of course, be anything between these two extremes.

Now because an electric field is invisible, the direction of its polarization is not obvious—unless perhaps one is within sight of the aerial responsible for it. But by connecting an alternating voltage between the top and bottom ("Y") deflection plates of a cathode-ray tube, so as to set up a vertical alternating electric field between them, the direction of that field is shown as clearly as could be by the vertical trace on the face of the tube, as in Fig. 3(a). Connecting instead to the "X" plates gives a horizontal line, as at (b). Any intermediate direction can be obtained by turning the

Fig. 2. In a radio wave two things are alternating at the same time, at right angles to one another and to the direction of travel. It is customary to regard the direction of polarization as that of the electrie field. The magnetic field in this diagram is supposed to be at right angles to the paper.



tube around so that the direction between the plates in use is neither vertical nor horizontal.

But users of oscilloscopes will know full well that there is an alternative method of getting an oblique direction, without shifting the tube. If equal voltages are applied simultaneously to both pairs of plates the line appears at an angle of 45° , (c) or (d) according to which "X" plate is connected to the top "Y" plate. The vertical and horizontal fields combine to cause a single diagonal field. And conversely this—or any other—single field can for purposes of calculation be imagined as split up into two fields at right angles to one another—resolved into its orthogonal components, to put it technically. If one wants the angle to be something other than 45° , it can easily be done by making the vertical and horizontal components unequal.

In the same way, diagonally polarized radio waves could be produced by combining vertical and horizontal waves. But seeing that the same result can be achieved much more simply by mounting a single aerial at the appropriate angle, why should one want to? Well, supposing one wanted to vary the polarization, it might be easier to do it by an electrical control than by moving the aerial itself. To be honest, however, my only reason for introducing this idea was as an intermediate step in the argument. When I said that alternating voltages were applied *simultaneously* to the two pairs of deflectors, I meant to imply that they were in phase. But suppose next that we insert



Fig. 3. Oscilloscope traces when a single alternating voltage is connected in phase to one or both pairs of deflector plates.

Fig. 4. Traces when the same alternating voltage is connected to both pairs of deflection plates, but with a variable phase delay in the connection to one of them. —are elliptical. Even when the phase difference is exactly 90° (quarter of a cycle or wavelength) the polarization is not necessarily circular; it is only so if the two components are exactly equal. When they are unequal, the polarization is horizontally or vertically elliptical, as at Fig. 5(a) and (b). And having studied the matter so far you will probably be the first to guess that a combination of unequal components with other phase differences yields ellipses at all sorts of angles, such as Fig. 5(c) and (d).

The full possibilities have even yet not been seen. The reason is that at any frequency above a very few cycles per second the eye sees a complete trace on the oscilloscope screen, whereas in fact there is only a single spot moving around it. There are two ways it can move around—clockwise and anti-clockwise—and only when the frequency is very low indeed can these be distinguished. But the two directions exist, whatever the frequency; and so do these two varieties of any elliptical polarization of waves. The difference between the two depends on which component is delayed relative to the other.

Another thing to remember about the oscilloscope picture is that it represents no more than a crosssection of a radio wave. It is a very valuable repre-sentation, nevertheless, because it shows so clearly the significance of terms such as "linear polarization, "circular polarization," and "elliptical polarization." (By the way, the first two are often regarded as special cases of elliptical polarization.) But if now we are sure that all is clear so far, let us pass on from a wave marking time, as it were, in two dimensions, and give it the order "quick march"-at rather more than Army speed, say 186,000 miles per second. If the polarization is vertical, the oscillograph spot ceases to move up and down across the face of the tube and shoots out to meet us at that speed. Stepping smartly to one side to avoid it, we catch a glimpse of it broadside on; the combination of the up-and-down and the forward movement results in the familiar wave shape, Fig. 2. The spot is not now confined to a vertical line, but moves in a vertical plane; that is why the wave it represents is often called plane polarized (instead of linearly polarized), especially by optical characters.



a variable phase delay in the connection to one pair. When the delay is zero the results are as already shown in Fig. 3. Shifting the phase causes the diagonal line to open out into an ellipse, then into a circle, then into an ellipse sloping the other way, and so on, as shown in the examples in Fig. 4, which are taken at intervals of 45° phase difference. At 360° , or one whole cycle, we are back at the starting point and the whole thing begins again

This shows us that if we combined two sets of radio waves, one polarized vertically and the other horizontally, we could get other kinds of polarization than diagonal by introducing a phase difference between them. It should surprise nobody to learn that the kind of polarization that results when the phase difference is 90° (or any odd multiple thereof) is called circular, and all the others—except the special cases at 0°, 180°, etc., which are called linear Of course a radio wave doesn't consist of a travelling spot; the spot is just something to help us to visualize a rapidly moving wavelike pattern of electric field. We can (I hope) easily see that a circularly polarized wave is represented by a corkscrew path moving forward in such a way that it traces out a cylinder. If we can't see this, then perhaps we had better take the clothes line out into the back garden and turn one end rapidly round in a circle at right angles to its length. Then, if we can still face the comments of the family and/or the neighbours, we might care to demonstrate the flattened corkscrewcylinder of elliptical polarization.

But what, you may ask, is the practical outcome of all this? The wayfaring man, though a fool, can see that some short-wave radio aerials are horizontal and some are vertical. If not quite a fool, he might wonder why. We, with our knowledge of radio and



Fig. 5. Traces resulting from unequal voltages; in (a) and (b) they are 90° out of phase, and in (c) and (d) have other phase differences.

now of polarization, may be quick to guess one reason, especially if we are familiar with the optical analogy of crossed Nicol prisms or Polaroid sheets. Reception of a vertically polarized wave is a maximum when the receiving dipole is vertical, and decreases to zero as it is turned horizontal. So it would appear that this supplies a method of completely separating two transmissions on the same wavelength. In practice it doesn't usually work out quite perfectly, but it does give a very useful degree of discrimination, which presumably is why the B.B.C. uses both vertical and horizontal on the same television channels.

The reason why discrimination is seldom perfect, and the reason why for some purposes vertical polarization is chosen and for others horizontal, are bound up together, and are due to the fact (familiar to optical students) that polarization is affected by the things that happen to it on the journey of life-transmission, reflection, refraction, and so on. For instance, in some circumstances vertically polarized waves are reflected less from the surface of the sea than are horizontally polarized waves. One of the troubles in certain kinds of radar is that the "targets" one wants to detect are liable to be concealed by reflections from the sea. This is the sort of thing that must be taken into acount when choosing which polarization to adopt for any particular system. But it is not always easy to make a quick decision, for the influence of polarization on wave behaviour is a very complicated subject in practice, necessitating a vast amount of experiment under working conditions.

Just now there is a good deal of interest in adopting circular polarization instead of linear, and I'm going to use the rest of the space explaining why.

The fact that an aerial arranged for, say, horizontal polarization is "blind" to vertically polarized waves, which is useful if one wants to separate two transmissions on the same frequency, is a disadvantage if one wants to receive signals on that frequency whatever their angle of polarization. True, an aerial at 45° would be equally responsive to horizontal and vertical waves—the loss as compared with a perfectly aligned aerial would be, as you have probably already worked out, nearly 30 per cent, or 3dB—but it would be totally unresponsive to waves polarized at 135°. You might think that nobody would be so silly as to radiate waves at 135°, but what about banking aircraft? Apart from that, the vicissitudes of travel sometimes (as I have mentioned) play tricks with the polarization, so that a wave starting off at 90° may be nearer 135° by the time it arrives.

If the aerial were made in two parts, one vertical and the other horizontal, both joined to the receiver in phase, it would be equivalent (as our oscilloscope showed) to a single diagonal aerial, so we would be no better off. But the use of a two-part aerial is on the right lines, for we have only to introduce a 90° phase delay in the connection from one of the parts to give the receiving system a circularly polarized



Fig. 6. (a) and (b) represent respectively the horizontal and vertical radiations of a circularly polarizing aerial system; (c) and (d) or (e) represent what is received by a similar aerial system after the signal picked up by the vertical portion is advanced or delayed relative to that from the horizontal portion.

characteristic. It is then equally responsive to waves linearly polarized at any angle; there are no blind angles.

The oscilloscope experiments should have made clear how the 90° phase difference between the outputs of the two aerial elements should give these results, but there is no harm in thinking of it a little differently. The reason why the straightforwardly connected outputs of a horizontal and a vertical aerial combine to give diagonal polarization is that the outputs, being in phase, are always present together and combine just like forces or winds or velocities acting in different directions. But when one is 90° out of phase with the other, its maximum occurs when the other is zero and cannot affect it; similarly the other can have its undisturbed say at least twice per cycle. Now, whatever the angle of the incoming wave may be, it can't be totally without horizontal and vertical. So some of it is bound to be received.

Now that we have thoroughly convinced ourselves that a circularly polarized receiving aerial yields an output from waves linearly polarized at any angle, we may be unprepared for the news that there is one kind of polarization to which it is blind. But there is no contradiction; the exception is not *linear* polarization, but circular polarization of the opposite direction of rotation. To prevent the mental haze from rolling in again it may be advisable to stop a minute and think this one out.

In doing so, we must be careful what kind of diagram we use; if we were to try vector diagrams, in which differences in phase (involving *time*) are represented on paper by differences in angular *position*, the chances of getting muddled in this

problem, which involves differences of both phase and angular position, would amount almost to a certainty. But I think there will be no danger if we regard the vertical and horizontal transmitting and receiving aerials as two independent channels of communication, in which (under ideal conditions) the vertical receiving aerial responds only to the vertical sending aerial, and similarly for the horizontal ones.

Fig. 6 (a) represents a brief sample of the radiation from the horizontal transmitting aerial. In order to cause circular polarization, the radiation from the vertical aerial must be 90° different in phase, so (b) shows what is going on there at the same time if the output is delayed 90° .

Supposing, for the sake of simplicity, that the receiving aerial is exactly a whole number of wavelengths distant, (c) shows what is being received by the horizontal part; it has the same phase as (a). That being so, what is received by the vertical part will have the same phase as (b). But because this is a circularly polarized receiving aerial system, it is provided with a 90° phase shift between its two parts. If the vertical part is advanced 90° relative to the horizontal (in practice this would be done by delaying the horizontal) the result is shown at (d), and this exactly reinforces what is being received on the horizontal part (c). But if the phase delay happened to be in the vertical part, the result would be as at (e), which would exactly cancel out (c) and give no reception.



Fig. 7. (a) End view and (b) side view of section of circular waveguide containing dielectric plate P for converting vertical (or horizontal) linear polarization to circular, or vice versa. The ends of the plate are fish-tail shaped to avoid impedance mismatching which would cause part of the wave to be reflected.

I have gone into this in some detail, not so much to show how one can very ingeniously invent a receiving aerial that doesn't receive, but to lead up to another use for circularly polarized waves. One of the biggest problems in radar is to distinguish one thing from another, especially when they are both at the same place at the same time. For example, aircraft to be located may be in the midst of a heavy fall of rain which entirely blots them out on the receiving screen by sending back a permanent echo. Can the rain echo be eliminated without eliminating the aircraft echo?

One feature that distinguishes raindrops from aircraft is their shape. A spherical drop, being symmetrical around its centre, behaves equally to the horizontal and vertical components of a circularly polarized wave, so the wave reflected from it is also circularly polarized. But in the reversal of direction, the rotation of the polarization is also reversed; so if (as is usual) the aerial that sent out the original wave is also used for receiving the echo, it is blind to the echo. An irregularly shaped body such as an aircraft, however, (or even a flying saucer edge-on) is practically certain to reflect a little more at one angle than at another, the result being that the echo wave is elliptically polarized and does not entirely cancel out at the receiver.

This idea is not particularly new; it goes back to somewhere near the end of the war, and various trials then and since have shown a reduction in rain echoes of up to 26 dB, whereas the echoes from "targets" are only 4-8 dB weaker. Provided that the target echoes are strong enough to stand this weakening and still be visible, the result amounts to a practically complete victory over the rain.

There are other uses for circular polarization; for example, radar and other microwave aerials usually have to be capable of being turned around freely, and if it is not convenient for the whole station to turn with them this means that somewhere along the waveguide feeding the aerial there must be a rotating joint. If the waves carried by the guide were circular throughout, this would hardly be a problem. But there are reasons for preferring rectangular (linearly polarized) waves, and if nothing were done about it the transmission would vary from a maximum to zero as the aerial system turned through 90°. The solution to the problem is to convert the linear wave into a circular one when it comes to the joint, and then back into a linear wave directly afterward.

And how, you may ask, does one convert linear into circular polarization and vice versa? In the circular-polarization aerial systems we considered earlier we assumed that two linear polarizations (horizontal and vertical) were laid on, but here in the wave guide there is only one. And even when we have two, how is one of them delayed 90°?

As regards the first point, we have already noted that any one linearly polarized lot of waves is absolutely indistinguishable from two in-phase lots, polarized 90° apart. For instance, one lot polarized at 45° may be, for all anyone can tell, really two lots polarized at 0° and 90°. (A south-west wind can be regarded as a south wind and a west wind blowing at the same time.)

If one had the two linear polarizations in different feeders, the phase difference could easily be introduced by making one feeder quarter of a wavelength longer than the other. The only snag would be if one wanted to work over a band of wavelengths.

Where there is only one feeder, conveying what might be a single lot of linearly polarized waves, the trick is to insert in its path something that slows waves of one polarization more than another. A simple something of this kind is a thin sheet of dielectric such as polystyrene. Fig. 7(a) is an end view and (b) a side view of a piece of circular waveguide with such a plate (P) in position. Let us suppose that the wave going in is vertically polarized. Then it can be regarded as consisting of two equal components, one parallel to the plate and the other at right angles to it. The former is slowed and the latter not, so if the length of the plate is designed to cause the parallel component to fall quarter of a wavelength out of step, the outgoing wave is circularly polarized.

A second plate at right angles to the first delays the other component so that it comes into step again and the polarization is vertical once more. By setting it at any other angle, the polarization can be made linear at some other angle. In the rotating joint the plate, being fixed in the rotating part, automatically turns to the correct angle for setting the linear polarization in line with the remaining run of rectangular guide.
Electronic Digital Computers

1. Basic Arithmetic Circuits

LTHOUGH an electronic computer forms an impressively complex device when viewed as a wholet, the circuits used are in many cases simpler than those of a superhet receiver or a television time base. This simplicity is essential to the success of a computing machine in many applications. The machine may, for example, form an essential part of an industrial concern's wages and accounting organization, and for such work it must have much greater reliability than apparatus intended for entertainment. Generally it consists of five hundred or more valve circuits linked together into a working whole, and the failure of any one part can put the entire machine out of operation.

To achieve this reliability it is common practice for individual circuit "bricks" to be developed and used over and over again in their original form. Connection of the "bricks" to other circuits is done via isolating amplifiers and cathode followers to prevent any possibility of the circuits they are feeding from affecting their stability. Opportunistic juggling of special characteristics of the same circuit element when used in different parts of an apparatus, though quite legitimate in a television set or reflex superhet, has to be kept to a minimum in the computing machine. It distracts from the margin of

British Tabulating Machine Company.
For a broad introduction to the subject see "Computors" by "Cathode Ray," Wireless World, October, 1951.



UNSET (b)

Fig. I. (a) Eccles-Jordan trigger circuit, used for various purposes in digital computers, and (b) its functional diagram.

By W. WOODS-HILL*

safety of the circuit as originally designed and increases the variable factors to a point where servicing becomes difficult if not impossible.

Before coming on to the actual arithmetic circuits of the computer it is necessary to look at some of the "bricks" from which they are composed. The two main ones are (a) a bi-stable element, usually an Eccles-Jordan' relay or trigger circuit and (b) a gate circuit. The Eccles-Jordan trigger is well known, but there are one or two points about it worth noting. In the circuit of Fig. 1 if V1 (a)



Fig. 2. Pentode gate circuit.

is conducting then its anode voltage has fallen from 135 V to 45 V and the grid voltage has risen from 30 V to +2 V, measured at the circuit end of the grid stopper. Exactly the converse has happened to triode (b). The source impedance of the anode circuit is low but the grid impedance is high.

When triode V 1 (a) is conducting the circuit is said by convention to be "unset" or "off". A negative pulse applied to the terminal marked SET will drive the V I (a) grid negative, effectively cutting it off, and the circuit will be unbalanced into the other state where V1(a) is cut off and V1(b) conducting. In this condition the circuit is said to be "set" or "on" and will maintain this condition indefinitely as long as the power is connected. Applying a negative pulse to the terminal marked UNSET will cut off V1(b) and restore the circuit to its original "off" state. The functional diagram for this trigger circuit is shown in Fig. 1 (b).

The gate circuit is not so well known and has many forms, but the simplest and easiest to design, though by no means the cheapest, is the pentode gate shown in Fig. 2. This is maintained cut off by connecting the grid through R_1 to -20 V. The suppressor "control voltage" is held at about earth potential. If positive pulses of 20 V or greater are applied to the grid through C1, the valve will act as an amplifier inverter and the pulses will appear in the anode with a peak-to-peak amplitude of about



Fig. 3. Diode gate circuit (a) with functional diagram (b) of its use as a logical " AND " element.



110 V. If the suppressor voltage is now brought down to -30V the electron stream will be prevented from reaching the anode and the amplified grid waveform will no longer appear there. This is a very attractive way of controlling a pulse stream, because the circuit is extremely tolerant to varying parameters. For example, the grid waveform can vary any amount from 20V amplitude upwards and the suppressor voltage anywhere from -30 V downwards, short of insulation breakdown, without interfering with the proper gating action.

If a gate circuit can be tolerated which requires the upper and lower limits of the voltage swings to be well defined then the diode circuit of Fig. 3 will cost considerably less. The output lead, though connected to h.t. via R_1 , will not rise more than a few volts above earth because both diodes are conducting. If a positive square wave is applied to one diode, say D_1 , this produces no change because the diode will cut off, as its cathode is driven more positive than its anode (negative anode volts). If. on the other hand, a positive pulse is applied simultaneously to both cathodes, then the output lead voltage will rise to the level of the highest square wave, because neither of the diodes will be conducting for the duration of the pulse. D₂ has effectively controlled the waveform from D₁. With germanium or selenium diodes the square waves are of 25 V-45 V amplitude and must be derived from

a low impedance source such as a cathode follower or pulse transformer to minimize breakthrough via the finite $(500 \text{ k}\Omega)$ back resistance or capacitance of the diodes.

The diode circuit in Fig. 3 illustrates another use of the gate-as a "logical" element which produces an output pulse only when two pulses are applied simultaneously to its inputs. In this sense it is known as an "AND" gate (because it needs a signal at one input and another signal at the other in order to operate) and the corresponding functional diagram is shown in Fig. 3 (b). Other versions with different methods of operation are "OR" and "NOT " gates.

In most computers the arithmetic is performed in binary notation. This is a "scale of two" notation, in which there can be only two possible symbols used as against the ten in conventional arithmetic. The carry, or point at which the same symbols are used over again, occurs after the tenth in decimals but after the second symbol in binary. The effect of this is such that one added to one (1 + 1 = 2)causes a carry to occur because it is in effect adding half the radix value[†] to half the radix value. In decimals 5 is half the radix value, and 5 + 5 is written $1 \leftarrow 0$, or more conventionally 10. In the case of binary notation 1 is half the radix, so that 1 + 1 = $1 \leftarrow 0$ or 10, which has a value of two decimal units.

It follows that binary two (or 10) plus binary two (10) will produce four as 100, four plus four will give eight as 1000, and so on, each binary place having double the value of the previous one. Thus we have:---

Decimal	0	I	2	4	8	16
Binary	00000	00001	01000	00100	01000	10000

Odd numbers are expressed as combinations of the above; for example, 5 in binary is produced by adding the symbol for four, 00100, to the symbol for one, 00001, and this gives 00101.

Because only two symbols are needed in binary notation they can be electrically represented in a computer as a voltage, say -100 V, being either present or absent at some point in a circuit, or by a switch being either closed or open, or by any other twostate device such as a trigger or a gate. The next thing to consider is how such electrical equivalents of binary notation can be manipulated by electronic circuits to perform arithmetic operations. To begin with take the simplest case-addition. Here, the circuit will have to obey the following rules of binary addition:-

Condition A	0 + 0 =	0
Condition B	1 + 0 =	1
Condition C	0 + 1 =	1
Condition D	1 + 1 = 1	$\leftarrow 0$ (or

All that is necessary, then, is to devise a circuit that will give outputs corresponding to the right-hand column of answers when inputs corresponding to the two left-hand columns of numbers to be added are applied to it. We may not be able to follow the arithmetic operations of such a circuit in the same way as we can, for example, the operations of a decimal adding machine, but that is simply because our minds are more accustomed to decimal notation. As long as the circuit obeys the binary rules, however, we can be sure that it will produce the right answers, and that is all that really matters.

The radix value is the number on which the system of numeration is based. In the decimal system it is 10, in the binary system 2.

· 10)

There are a number of different types of adding circuits, but a simple one that will give the answers for conditions A, B and C and partly for D is shown in Fig. 4. If by convention a binary digit (that is, a 1) is represented by a fall in voltage of 100 V, say from + 150 V to + 50 V, then condition A (0 + 0 =0) means that the voltage at both x and y is +150 V and both halves of V1 are conducting, because both grids are going positive with respect to earth. This results in no output from the anodes of V2 (which is actually an inverter to counteract the phase reversal produced by V1). Condition B (1 + 0) means that the voltage at x is low while y is still high, so the junction of R_1 and R_2 falls approximately 50 V. This results in the grid of triode (b) going negative because it is connected low down the voltage-divider chain R₃R₄. The grid of triode (a) is still positive because it is connected higher up the chain. That is, (a) is conducting and (b) is cut off. After inversion by V2 this results in a fall of voltage (representing a 1) from the "sum" output (V2(b) conducting) but nothing from the "carry" output. Condition C produces the same result as B because the junction of P and P again for the formation of the same result as $\frac{1}{2}$ for $\frac{1}{2}$ and $\frac{1}{2}$ of R1 and R2 again falls 50 V as before.

Condition D (1 + 1) means low voltage on both x and y, which caused a fall of 100 V at the junction of R_1 and R_2 . Section V1(b) is driven even farther negative with the same result as in conditions B and C, but V1(a) grid is now driven below cut-off for the first time and its anode voltage rises. The output after inversion now shows a digit (fall in voltage) on both lines, which is not correct. Only a "carry" digit should be present $(1 + 1 = 1 \leftarrow 0)$ and V2(b) should not be conducting. A further valve must therefore be added to suppress the "sum" output for condition D. This takes the form of V3 in Fig. 5 which shows the complete adding circuit. The grid of V3 is fed from the anode of V1(a) which, as just described, is low under all conditions except the last, so that normally V3 is cut off and has no effect; but under conduct, thereby



Fig. 6. Functional diagrams of (a) half adder, (b) complete adder with (c) its abbreviated form, and (d) an adder with a storage device for holding the "carry" output.

shunting the anode of V1(b) and effectively suppressing the "sum" output.

The type of circuit just described has become known in the computing world as a "half adder" and can be symbolized as in Fig. 6(a). The name springs from the fact that it will perform the addition of two binary numbers quite correctly but does not provide a means of accepting a "carry" produced by the addition of lower digits. For example, when one goes through the motions of adding 25 + 26like this:—

$$+\frac{25}{51}$$

one says "six plus five equals eleven; one down, and carry one." Then, "two

carry one." Then, "two plus two, *plus one carried*, is 5." Not only has 2 + 2been added to form a temporary answer of 4 but a second addition must be immediately performed to add the one and clear the "carry." Quite logically, then, a second identical circuit can be included in series with the first "half adder" to cope with the third input for the incoming "carry" digit. The symbolic circuit for an adder then becomes Fig. 4(b) with z as the third input, and this in turn can be abbreviated to Fig. 6(c).

The explanation will not be clear until it is realized that this adder will deal with, say, two 10-digit binary numbers presented simultaneously to the two inputs x and y in pairs in serial form, starting with the lowest value digit. For example, 13 + 13 would look like



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1101 + 1101 and would be presented to the x and y inputs in the form shown in Fig. 6 (d). As each pair cf digits is presented the addition may or may not form a "carry," but if one is formed it must be stored until the next higher value digit pair is presented and then added to the sum of this next pair via input z. The storage, represented by the device D in Fig. 6(d), need consist of nothing more complicated than a lumped-inductance delay line with a delay time equal to one digit period, or a trigger circuit (similar to Fig. 1) which would be "set" by the carry pulse.

In most digital computers the numbers to be added are held in a storage system, say a magnetic drum, and they emerge in some kind of sequence as the drum rotates. In the ordinary way, then, the two numbers in Fig. 6(d) would not arrive simultaneously at the adding circuit. To achieve this type of presentation it is necessary to use an additional temporary store which will hold the first number until the second one arrives and then feed them both, a pair of digits at a time, into the adder. Such a storage system is usually known as a " shifting register."

If, as depicted in Fig. 7, ten trigger circuits are set up in a row they will store (by their states of operation) any discrete pattern representing a binary number of not more than ten digits. To get a number to shift along this row requires the addition of two gates between each trigger and a source of negative pulses. Trigger 1 controls two gates G_1 and G_2 such that G_1 is open and G_2 closed when T_1 is "set" and the reverse if T_1 is "unset." The gates are connected to the next trigger T_2 so that a pulse from G_1 will "set" T_2 while a pulse through G_2 will "unset" it.

Because T_1 governs which gate is open, then on the application of a negative pulse to both gates only one gets through and causes T_2 to assume whatever state T_1 was in at the instant of applying the pulse. Similarly T_2 influences T_3 , T_4 , and so on up to T_{10} . Thus the application of one negative pulse to all the gates will cause whatever pattern was in the register to shift one stage along. Applying ten pulses will shift the pattern right off the end unless the end pair of gates is connected back to the first stage, in which case the pattern will have recirculated back to its starting position.

It is now possible to assemble in one block diagram (Fig. 8) all components needed to perform the addition of two binary numbers. The sequence of events is as follows. Trigger T_P is continuously triggered by pulses derived from an oscillator. The resulting pulses from T_P are applied to a gate. A "start" pulse "sets" T_0 which opens the gate. The output of the gate causes both registers to shift their contents





into the adder. The adder then sums each pair of digits and takes care of the "carrys" as it goes, returning the sum to register B. Notice that register B loses its initial contents (if any) which are replaced by the sum, but on the other hand register A has a feedback loop so that not only does it present its digits to the adder but preserves them for future use. At the end of ten pulses T_G receives a "stop" pulse, becomes "unset", closes the gate, and stops the action just as the digits in register A have returned to their initial positions and the sum is complete in register B.

The same equipment can be used to do multiplication. To multiply, for example, $3 \times 4 = 12$ the binary equivalent of 3 is inserted in register A, zero into register B, and the circuit is made to perform four addition cycles. The 3 will then be added four times to the contents of register B (initially zero), which gives:

$$0 + 3 + 3 + 3 + 3 = 12$$

How the machine causes the value of the multiplier factor to control the addition cycles will be explained later.

Before continuing it may be of interest to look at the practical circuitry of the part of Fig. 8 comprising triggers T_0 and T_P and the gate. In Fig. 9, trigger T_0 (V1) is providing a controlling voltage from its grid circuit via a cathode follower, V2, to the suppressor of gate V3. The grid of V3 is supplied with a train of square waves derived from trigger T_P (V4). It is

assumed that V4 is being triggered by some externally generated stream of pulses at, say, 30 kc/s. As the grid swing of a trigger is -30 V to +2 V then if V1 is "set" the pentode gate V3 is open (suppressor at +2 V) and the amplified inverted pulses will appear in the pentode anode circuit; but if V1 is "unset" then the gate is closed because its suppressor is at -30 V. Notice that there is no isolating cathode follower between trigger V4 and gate V3 because the grid is the first element in the pentode electron stream and contains very little modulation which might accidently upset trigger V4.

To return to arithmetic operations, subtraction can be achieved, apart from building a subtractor, by using the adder to add numbers in complement. A complement is the difference between a number and its radix value. In decimals they are:—

Number	1	2	3	4	5	6	7	8	9	10
Complement	9	8	7	6	5	4	3	2	1	0

which is ten variations and would be quite difficult to produce automatically; but binary comes to the rescue once again because there can only be two:—

which in practice simply means inverting the voltage swings with a valve so that a falling voltage becomes a rising voltage.

Why adding a complement is tantamount to subtracting a true, will be easier to understand in decimals. Take for example the subtraction 6 - 3. The complement of 3 is 7, so if we add 7 to 6 we get $6 + 7 = 1 \leftarrow 3$. Ignoring the "carry" the answer is correctly 3.

It will now become clear how the cycles of multiplication mentioned above can be controlled by a number. If our example 3×4 is taken, then the figure 4 is placed in complement (6) into some counter whose capacity is exactly 10. At the end of each cycle of addition the counter is arranged to advance one step, so that at the end of the 4th cycle the counter will

Fig. 9. Circuit of system for producing a predetermined number of shift pulses in Fig. 8.



Fig. 10. Simplified block diagram of complete computer.

"spill over" (because 6 + 4 = 10) and this "spill over" can be used as a "stop pulse" in Fig. 8.

Division is very much the reverse of multiplication in that one factor is repeatedly subtracted from the other, and a tally of the number of times a subtraction is successfully achieved is accumulated in a counter. For example $8 \div 2$ is performed as 8 - 2 - 2 - 2 - 2, the answer being 4 subtractions of two. The action is stopped as soon as 8 is reduced to zero or, in the case of an imperfect divisor, becomes a negative value.

To describe all the other circuits of a digital computer would be outside the scope of this article, but it may be helpful to glance at the general organization of a typical machine. The block diagram Fig. 10 gives some idea of this. The whole of Fig. 8 and some other circuits for doing arithmetic operations have been condensed into the block marked "arithmetic unit." The other blocks contain equipment built from similar "bricks" but connected to perform different functions. First of all the "control unit." The job of this

First of all the "control unit." The job of this unit is to feed out to the whole of the rest of the machine trains of pulses and start/stop pulses as they may be required, also d.c. voltages to gates for routing numbers from one unit to another. It will activate these controls when instructed to do so by means of "program" numbers (instructions) delivered to it from a "program" section of the storage unit.



This storage unit consists of two sections, usually of a magnetic drum, capable of storing 2,000 or more groups of binary digits, each group (known as a "word") containing approximately 32 binary digits. One section stores the numbers upon which arithmetic is to be performed while the other is the "program" section devoted to storing instruction numbers. The storage system as a whole will be capable of sending numbers to the control unit, sending to and reaciving from the acithmetic





receiving from the arithmetic unit, and sending to and receiving from the input and output units.

These input and output units can take many forms, from magnetic tape to typewriters, and they contain electronic circuitry for coding decimal numbers to binary notation and decoding back again. Punchedcard reading and printing equipment is used a great deal, and a typical card for use with this is shown in Fig. 11. From the top to the bottom of the card there are ten possible places to punch a hole, 0 to 9, so that a hole punched three positions down will have a value 3. The column or position from the righthand edge of the card will signify its decimal value, i.e., 3 punched in the second column will have a value of 30. As very few numbers require more than ten columns, lines are drawn down the card to segregate each number, the sections between lines being known as fields. Fig. 11 shows 30 punched in the first field and 123 in the second. These cards, when passed

by means of rollers through the machine (usually at about 100 per minute), allow contacts to drop on to a sensing roller wherever a hole is punched. The impulses from these contacts are then fed via a coding device to the storage unit. The punched-card output unit consists of a printing machine in which 80 letters in a complete line of type can be printed simultaneously. This gives a rate of typing equivalent to 1,000 words a minute. Outgoing information from the storage unit feeds via a decoder to electromagnets controlling the setting of the type bars.

Finally, it is as well to recall that this article is an introduction to computer circuits in general and is not intended to represent any particular machine. Though it is quite feasible to build a computer with as few as 5 different types of circuit, most machines have 20 or more. The S.E.A.C. machine in the U.S.A. on the other hand is said to use only 3 types in its "logical" circuits.

VERY LOW FREQUENCIES FROM TAPE

IN the conventional magnetic tape reproducer the output voltage from the head is proportional to the rate of change of flux induced by the tape. Consequently, for a constant peak induction the output falls with frequency at 6 dB per octave and at very low frequencies the output may be little more than the hum and noise level of the amplifier. This is not likely to occur at audio frequencies but would preclude the use of magnetic tape for data recording which involves infra-sonic frequencies.

A special type of head has been evolved by E. D. Daniel (B.B.C.) for this type of work and is described in the *Proceedings of the Institution of Electrical*



Simplified diagram of flux-sensitive, low-frequency magnetic reproducing head. Institution of Electrical Engineers, Part B, No. 4, 1955). Vol. 102 (July The object of the design is to derive an output which is proportional to the magnitude and not to the rate of change of flux; the system can, in fact, be used to explore a transient slowly and in detail. The basic principle is to cause the signal flux to modulate а high-frequency flux (which does

quency hux (which does not, in fact, reach the gap) and to pick up the resultant product by means of a search coil. There are points of similarity with a secondharmonic magnetic modulator and with a sensitive magnetometer described by T. M. Palmer (*Proc. I.E.E.*, Part II, October 1953). In the accompanying simplified diagram it will be seen that the back leg of the magnetic core of the head encloses a single straight conductor, arranged symmetrically along the axis of a coil wound over the outside of the core. The latter is conveniently built up from pieces of one of the ferrite materials with grooves ground to fit on each side of the wire. A current of the order of amperes at a high frequency f in this wire generates a cylindrical flux inside the core, but as the linkage with the coil is small there is no significant output until a polarizing flux f_s due to the signal is applied in a direction at right angles to the high-frequency flux. The magnetic characteristics of the core material, which is being worked near saturation, will also no longer be symmetrical, and even-order harmonics, of which the second is the most important, will appear in the coil. After passing through a band-pass filter to select the second-term products, the output is fed to a balanced phase-sensitive detector in which the "carrier" frequency is balanced out by a variable-phase auxiliary oscillator (2f).

The head is sensitive to stray fields and must be well shielded. Any residual flux can be balanced by a d.c. current (of the order of microamperes) through the coil.

To take full advantage of the system the length of contact between tape and head should be at least half a wavelength of the lowest recorded frequency, but even with standard core shapes the published response shows no falling off below 1 c/s at a tape speed of 1 in/sec.

It is suggested that the system will find its chief application where low frequencies and long playing time are required as in cardiography or encephalography, when it may be necessary to keep observation of long periods for possible abnormalities.

Component Tolerances

By H. S. JEWITT, B.Sc.(Eng.)*

A Reminder of Their Effects on Circuit Performance

T is often thought that the function of the circuit design engineer is to produce a circuit that will meet a given specification. This is, of course, quite cor-rect, but one aspect of his job which tends to be overlooked is that he must also be what, for want of a better term, may be called a tolerance engineer. To produce in the laboratory a circuit to perform some operation may or may not be difficult according to the complexity of the required operation, but this is not enough: it is vital to ensure that the circuit can be reproduced in even small quantity production with an acceptable performance, as otherwise the design side of the work has been wasted. To this end, the effect of variation of the circuit components on the circuit performance must be examined, bearing in mind the fact that component manufacturers must, of necessity, be allowed some latitude in the values of their production components.

This examination must be practical in its approach. It is possible to specify highly accurate components in many places in a circuit and thereby produce a level of performance which may be unnecessarily high and will certainly be very costly. As an instance of this, the resistor may be considered. The ordinary composition resistor is a cheap component and is commonly available in tolerance ranges of $\pm 5\%$, $\pm 10\%$, and $\pm 20\%$, the price varying, of course, with the tolerance demanded. The wire-wound resistor and the high-stability carbon type are both available with better tolerances, $\pm 2\%$ or $\pm 1\%$ or even better: but these components are much more expensive than the composition resistor, and replacement of a large proportion of the composition resistors in an equipment by the high-accuracy types might well double or treble the cost of the equipment. It may be that the level of performance demanded is such that this has to be accepted, but this is a rare case, and it should be the aim of the tolerance examination to make certain, not only that the circuit performs satisfactorily when component tolerances are taken into account, but also that high-accuracy components are called for only when it is strictly necessary that they should be used. A very practical point about the use of such components which probably lies within the experience of most design engineers of today is that delivery dates tend to be months later than those for the lower-grade components.

Another aspect of tolerance engineering is to determine what the circuit performance will be with specified components. It is well known that, for example, a wire-wound resistor is not a practical resistor for use in an i.f. amplifier at, say, 60Mc/s centre frequency, and neither, generally, is a spiraltrack high-stability resistor. Thus, the designer of such a circuit must accept the fact that he has no

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option but to use composition resistors, and decide (a) whether his chosen circuit will give the desired performance with the tolerances of these components or (b) whether some other configuration would give a better performance with these tolerances.

It should be, but unfortunately is not always, realized that to find a resistor colour-coded 1000 ohms whose value is exactly 1000 ohms is a rarity, and that, for a $\pm 10\%$ tolerance, the value may lie anywhere between 900 and 1100 ohms. Nor is it correct to assume that the majority of a given batch of components will have values centred closely around

the nominal value, particularly for the larger tolerances: they tend, rather, to have values which lie in two bands towards the two edges of the permissible range of values for the given tolerance. The circuit component which is an exception to this is the valve; a given batch of valves tends to have parameters centred closely about some value in the permissible range of those parameters, the value usually differ-



Fig. 1. Typical biasing circuit to illustrate the effects of component tolerances.

ing for different batches of valves. Thus, it is most important to know what the expected performance variation with component tolerance will be.

Practical Example

A simple example of this might be a biasing chain as shown in Fig. 1. In this circuit, R_1 and R_2 form a potentiometer network across the 100-volt negative supply to provide the correct bias voltage for the valve grid. Let it be assumed that the valve anode current is to be cut off; in some possible circuits a very small flow of anode current could cause the valve power dissipation rating to be exceeded so that this requirement is not a fictitious one. If the valve data is consulted, it may be found that the nominal cut-off voltage is -10 volts. Then R₁ and R₂ may be chosen to produce a little more than this minimum value: the values chosen might well be $R_I = 5.6 \ k\Omega$, $R_2 = 47$ k Ω , and the resultant bias is then -10.6volts. But if these resistors are selected to a 20% tolerance, and R_1 is at its minimum value (= 4.48 k Ω) and R_2 at its maximum value (= 56.4 k Ω), then the bias might fall to 7.35 volts. It is very clear that the choice made, of a 20% tolerance, will lead to un-

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satisfactory performance in a large number of the sets built using this circuit. Reduction of the tolerance limit to 5% effects an improvement, and gives a possible minimum bias level of -9.73 volts. If the nominal bias level, that is, the level with both resistors at their nominal value, be now raised from 10.6 volts to a value of about 11 volts, the valve will always be cut off.

So far, attention has been focused on the potentiometer network, and it has been seen that the effect of resistor tolerance can be satisfactorily overcome. Now the valve itself must be considered. The manufacturer states that the cut-off bias is -10 volts for a nominal valve, and, in fact, if he is consulted, he will state that the possible spread of cut-off voltage for the valve type is from -7 to -13 volts (taking typical figures). In the light of this information, the whole problem has to be reconsidered, with a new minimum bias level of -13 volts as the target, requiring a nominal level of perhaps -14.5 volts. The alternative is, of course, to recognize that an equipment in which all parameters simultaneously assume their worst possible values will be rare, and that this equipment may be rejected: but this can be a dangerous practice, for it leads to a relaxation of standards and possibly to future servicing diffi-culties with the equipment. Whichever philosophy is adopted, these tolerance calculations must be done at a very early design stage if they are going to be This may be explained by slightly done at all. extending the example above. Let it be assumed that the valve is to be triggered by a positive trigger pulse which must drive the grid to zero grid voltage for satisfactory triggering. Then the original bias level requires a trigger pulse of 10 volts, and the final bias level one of 14.5 volts amplitude. Such a change in triggering pulse level is considerable: if the complete equipment had been designed and built without a consideration of tolerance effects, and it had been found, by experience, that this change was necessary to obtain a satisfactory performance, a complete redesign might well be needed to produce it.

Mathematical Approach

The example of tolerance effects considered above has been a simple one. It should be noted now that the use of the mathematical differential of the circuit equation will eliminate much laborious arithmetic, provided that it is realized that the differential approach only gives exact answers for infinitesimally small changes: but it is sufficiently accurate for this purpose over normal electronic engineering tolerances. Thus the bias voltage V_b of Fig. 1 may be expressed as

$$\mathbf{V}_b = \frac{\mathbf{R}_1}{\mathbf{R}_1 + \mathbf{R}_2} \ . \mathbf{V}$$

This expression may be differentiated with respect to R_1 , assuming R_2 to be constant, and then with respect to R_2 , R_1 assumed constant, to produce two equations relating the change of V_b to the changes of R_1 and R_2 . These equations are

$$\frac{dV_{b}}{dR_{1}} = \frac{R_{2}}{(R_{1} + R_{2})^{2}} V \quad (R_{2} = \text{constant})$$
$$\frac{dV_{b}}{dR_{2}} = -\frac{R_{1}}{(R_{1} + R_{2})^{2}} V \quad (R_{1} = \text{constant})$$

Then the change of V_b for a change dR_1 in R_1 is given by

$$dV_{b(R_1)} = \frac{R}{(R_1 + R_2)^2} V dR_1$$

and similarly for $dV_{b(R_2)}$, and, since interest here lies in the minimum value of V_{b} , this can be stated as

$$V_{b(m_{10})} = V_b - dV_{b(R1)} + dV_{b(R2)}$$

The positive sign for $dV_{b(R2)}$ is necessary since the sign of the differential is negative, indicating that V_b decreases as R_2 increases.

This simple example indicates both the need for, and the solution to, tolerance engineering of electronic circuitry, and, in general, this may be done, circuit by circuit, for the whole of an equipment. The process tends to become laborious for a large equipment, but it is hoped that the necessity for it has been adequately established. In more complex circuits, the same method may be used to great effect; the circuit equation is generally known from the design process, and differentiation of it is not generally difficult, whereas the arithmetic involved in considering the variation as applying to each component in turn with numerical values is often staggering. It is not proposed to give an example of this, as every case is a special one in this work, and no useful purpose could therefore be served.

Preset Adjustments?

From what has been said so far, it will be clear that one method of solving the problem of performance variation with component tolerance is the use of preset adjustments. Thus, in the example above, a suitable portion of R_1 and R_2 could be inserted in the form of a variable potentiometer, which could be set on each individual unit to give a suitable cut-off bias for the valve. This seems, at first glance, to eliminate the whole need for tolerance engineering, but this is a false conclusion. The preset itself has tolerances, and its value must be so calculated that its range of adjustment covers the tolerance variation of R_1 , R_2 , the cut-off bias voltage, and itself, so that tolerance examination is still necessary. The whole of R1 and R2 could be made a potentiometer to avoid tolerance calculations (the brute force attack) but this must give only a rough control over V_b , and may also need a large component to dissipate the power involved.

The biggest objection to this solution to the problem is that, remembering the number of circuits involved in almost any piece of equipment, the numbers of preset controls needed would be fantastic. It is probable that some presets must be used, and indeed should be used to provide a good level of performance, but the number of these must be cut to the bare minimum, on both cost and convenience grounds. The cost angle needs little elaboration-a variable resistor is more expensive than a fixed one. Convenience is a factor affecting the ultimate user of the equipment. Adjustment of presets is necessarily a job for a skilled man, who is neither always available nor cheap to employ, so that his attentions should be needed as little as possible. The user himself rapidly gets a poor opinion of a set needing frequent skilled attention. Last, but not by any means least, it is surprising how many people, whether skilled or unskilled, cannot resist the temptation, if they see a knob or other means of adjustment, to twiddle it

So, presets must be reduced to the minimum and preferably be hidden away, with screwdriver slots rather than knobs and possibly locking devices in addition. In the process of hiding them away, however, it is as well not to forget that they do need adjustment, and to arrange things so that the preset and the indicator by means of which it is being set can be effectively used by people of normal proportions, with their eyes in the normal places. Few things are quite so irritating to the maintenance engineer as finding that the only person able to adjust a given preset is a dwarf with four-jointed 8-ft arms and eyes on flexible stalks.

It is usually possible to select the value and position

Modified A.F. Millivoltmeter

HIS instrument is based on the design for an audio-frequency valve voltmeter by S. Kelly⁺, and incorporates additional features to meet specific requirements.

An instrument was first built to the published design, incorporating a meter having a full-scale deflection of 1 mA, with a copper-oxide bridge rectifier and a suitable series resistor. This performed quite well, but it soon became apparent that operation of the range switch was accompanied by switching transients in the meter.

Consideration showed this to be due to the fact that the range selection potential divider acts as such, not only to a.c., but also to the standing cathode current of the first valve. When changing, for example, from the $\times 1$ to the $\times 0.1$ positions, there is a change of about 30 volts in the d.c. potential at the moving arm of the switch, and, owing to the good l.f. response of the amplifier, this appears as a surge at the output.

In the modified design the cathode resistor is separated from the potential divider, feeding the latter through a large capacitor which blocks the d.c. but offers low impedance to a.f. No alterations to the values of the resistors are necessary. The cathode resistor may be of the order of $50 \ k\Omega$, and a high-stability type is unnecessary. The effective impedance of the cathode-follower load is approximately halved by this arrangement, thereby reducing the maximum signal which can be handled without distortion, but this will not usually be a material factor.

Some difficulty was experienced with the original design in securing a linear scale. This is, of course, a normal feature of rectifier-type voltmeters, due to the characteristics of metal rectifiers. The meter which it was desired to employ had a resistance of 50 ohms and a full-scale deflection of 1 mA. A curve was plotted (Fig. 1) showing the relationship between scale reading and applied voltage when used with an instrument-type bridge rectifier and a series

* Metropolitan Relays Ltd.

+ Wireless World, June 1951.

WIRELESS WORLD, NOVEMBER 1955

in the circuit of a preset control so that it compensates for the effect of a number of component tolerances. Before presets are positioned in the circuit, therefore, their effects should be carefully considered to see whether their number can be reduced by altering their placing in the circuit.

It is hoped that this article will underline the need for tolerance engineering at a very early stage in the design process. The importance of this, and of its solution in terms of adequate performance and reduced cost, cannot be over-estimated in an industry in which ever-increasing circuit complexity poses new problems of economic manufacture and satisfactory use of equipment.

> Reduced Switching Transients and a More Linear Scale By R. SELBY*

resistor of 3,800 ohms. A further curve was taken with the same meter and a pair of germanium diodes in a half-bridge circuit as shown in Fig. 2. This curve (not reproduced) showed a higher degree of linearity owing largely to the higher current being drawn with this circuit.

A linear scale is obtained in commercial multi-





purpose test meters, but only at the expense of lower resistance and higher consumption than on the corresponding d.c. range. For example, the latest model of a very widely used instrument consumes 10 mA at f.s.d. on the 10-V a.c. range and 40 mA on the 2.5-V range, whilst another model consumes 5 mA at 10 volts. The use of such meters externally as the indicating device in the original valve voltmeter circuit imposes certain restrictions on accuracy. At low frequencies the reactance of the coupling condenser becomes appreciable in relation to the meter resistance, whilst the impedance of the load presented to the cathode follower is so low as to reduce seriously the undistorted output.

One method of overcoming this difficulty is to employ a low-consumption meter with a specially calibrated non-linear scale. This course did not appeal to the writer who is more at home with a soldering iron than a draughtsman's pen. In the original version it was possible to overcome the difficulty to a reasonable extent by feeding the meter through a fairly high series resistor, so that variations in rectifier resistance represent only a small fraction of the total resistance in circuit. Clearly, sensitivity is lost. In many designs, the meter is fed from a pentode or high-impedance triode in order to improve linearity.

In the modified design, the cathode-follower output has been retained, but the meter has been placed in the negative feedback circuit, thereby substantially linearizing the scale shape. It was found that with this system there was no visible difference in linearity between the copper-oxide full bridge and the germanium half bridge types of rectifiers.

A further difficulty arose when attempts were made to modify the ranges. For the writer's purpose, the desired scales ranged from 0-50 mV to 0-5 V only, and it was hoped that this reduction in sensitivity, together with improved long-term stability, could be achieved by increasing the percentage of negative feedback. Unfortunately this was found to cause serious instability, the meter swinging violently at a periodicity of 2 or 3 c/s.

The circuit was therefore finally modified to that



Fig. 3. Modified circuit in its final form. Components are numbered to conform with the original circuit (June 1951).

List of Components R_1 2.2MΩ R₁₆ 390kΩ C_3 C_4 C_5 C_6 C_7 C_9 $16 + 16 \mu F$ R_2 3.3kΩ R₁₇ 0.5µF $100k\Omega$ R₁₈ 50k Wire-wound R_7 680Ω $0.5\mu F$ R_8 470kΩ R20 10kΩ Wire-wound $0.5 \mu F$ R₉ 570kΩ R₂₃ $10k\Omega$ Wire-wound 2μĖ R₁₀ $100k\Omega$ R₃₃ 890kΩ High-stability $16\mu F$ R₁₁ 470Ω Wire-wound 50Ω Variable R₃₄ $47k\Omega$ C11 8µF Metallized paper (or "Super-R₁₃ R36 3.8kΩ High-stability lytic") R₁₄ $20k\Omega$ C_{13} 8µF (T.C.C. " Superlytic ") $R_{1\delta}$ 100kΩ C.2 0.01µF vi, v2, ٧3 6AM6 or EF91

shown in Fig. 3, from which it will be seen that the amplifier (V3) is now direct-coupled to the cathodefollower output stage, whilst a direct-coupled feedback path is now provided through R_{33} from the output valve cathode to the cathode of V2. Two sources of phase variations are thus removed and a very high feedback factor may be used without trouble, if required.

A second feedback path is provided by returning the meter also to the cathode circuit of V2. Apart from the normal overall improvement to the amplifier, this has two beneficial effects. The reactance of the coupling condenser C₁₁ no longer introduces error at low frequencies and the meter scale becomes substantially linear, because variation in rectifier resistance introduces a compensatory variation in the gain of the amplifier. It was found, however, that there is a limit to the amount of feedback which can be used in this path without instability. This again showed itself as a surging or oscillation at very low frequency. The procedure in choosing component values is, therefore, to apply the maximum feedback in the meter circuit compatible with stability, after which selection of R₃₃ may be made in accordance with the overall sensitivity desired. This will be determined by the ranges required and the particular meter used. The resistor values quoted relate to the meter and ranges referred to earlier. It should be noted that although the lowest range has been raised to 0-50 mV, this does not imply that the highest may be 0-50 V, since such a signal would overdrive the input cathode follower.

A few other small modifications should perhaps be explained. The cathode bias resistors of V2 and V3 have higher values than the 100 ohms shown in the original circuit. Using 6AM6 or EF91 type valves, 100 ohms gives only about 0.3 V bias. Further, the use of higher values, unbypassed, introduces some secondary feedback within the main loop, thereby improving stability. The screen feed resistor of V3 has been reduced from $470 k\Omega$ to $390 k\Omega$ in order to lower the anode potential of V3 and therefore the grid potential on V4. The suppressors of V1 and V4 have been strapped to the cathodes rather than the anodes, to conform with the valve makers' recommendations. The use of $0.5 \mu F$ condensers for coupling and decoupling in place of $0.25 \mu F$ has no significance —a quantity of $0.5 \mu F$ condensers were available.

The grid resistor of V2 is returned to earth instead of to the lower end of R_{11} as in the original circuit, since it was found that this was one cause of the meter surging during range switching. This is due to the removal of feedback during the momentary break in the circuit when the switch arm is moving from one contact to the next. Hum and signal leakage (by stray capacitance) are temporarily amplified with the full gain of the amplifying stages. This can be seen more easily by re-drawing the essentials of this part of the circuit as in Fig. 4. This trouble would not arise with a switch which shorted adjacent contacts during rotation. It is desirable to screen C₄ and its associated wiring to reduce signal leakage by stray capacitance, when a non-shorting type of switch is used.

Switch is used. The condenser C_{13} coupling the cathode of V1 to the potential divider should be of large capacitance but small physically. A T.C.C. "Superlytic" 8μ F, 150-V working type was used, and it was found possible to dispense with C₄, provided one remem-



Fig. 4. Simplified detail of grid-cathode circuit of V2 to show possibility of hum and signal leakage amplification when the range switch breaks contact between adjacer positions.

bered to leave the range switch on its least sensitive position for a minute or so after first switching on from cold, in order to allow polarization to become complete. This type of capacitor has an insulation resistance, after polarization, approaching that of paper types. However, as at the time of writing it is not yet widely available, a normal electrolytic may have to be used in many cases, when C_4 must be retained to isolate the grid of V2 from d.c. leakage. The calibration checking arrangements given in the original description may be retained, but have been omitted from the diagram for simplicity, as has the power supply section.

One desirable feature not incorporated would be automatic protection against accidental meter overload. This might possibly be achieved to some extent by reducing the resistance of R_{23} to a figure which would cause the output to flatten off slightly above that required for full-scale deflection of the meter.

Television Interference

DURING some experimental work on a two-band television tuner it was found that some interference with Channel 1 was being experienced from Channel 9. The set had the new standard intermediate frequencies of 34.65 Mc/s vision and 38.15 Mc/s sound.

For Channel 1 the oscillator was at 45+34.65 = 79.65 Mc/s. Its second harmonic was thus 159.3 Mc/s. The difference frequency from Channel 9 was 194.75 – 159.3 = 35.45 Mc/s, which was within the vision channel i.f. pass band of 34.65-37.15 Mc/s. It thus produced a beat of 35.45 – 34.65 = 0.8 Mc/s with the Channel 1 intermediate frequency and a corresponding bar pattern on the picture.

This effect was experienced in a district near Alexandra Palace where the Band III signal is relatively weak. It would obviously be more likely to occur near the Channel 9 transmitter where the conditions would be reversed. It should be emphasized that the equipment used was experimental and that no difficulty was found in overcoming the interference. At the time an r.f. coupling circuit effective in both bands was being used as a temporary measure, so that the tuner had only one effective signal circuit to discriminate against the effects. Most commercial sets have tuners with three such circuits and conditions would have to be very unfavourable for it to occur with these.

Manufacturers' Products

NEW EQUIPMENT AND ACCESSORIES FOR RADIO AND ELECTRONICS

New Single-sideband Receiver

THE receiver illustrated is the Racal Type RA9 which has been developed in conjunction with the R.A.E., Farnborough, for ground use in ground-to-air single-sideband radio-telephone communication in the frequency band 2.5 to 20 Mc/s.

Frequency control of the carrier re-insertion oscillator is facilitated, first, by employing a temperature-stabilized crystal oscillator and, secondly, by not entirely suppressing the carrier, but radiating one with an amplitude inversely proportional to the average depth of modulation. Thus in the absence of modulation the carrier is radiated at maximum transmitter power and reduced to a low level with full modulation. This system provides a strong, but intermittent, signal for operation of the a.f.c. circuits in the receiver.

The advantages claimed for single-sideband operation are: improved signal/noise ratio (about 9 db), freedom from selective fading and cross modulation and more economical operation of the transmitter. Some features of this particular receiver are: sensitivity $1 \,\mu V$ input for 25 db signal/noise ratio up to 10 Mc/s and 20 db over; unwanted sideband rejection better than 55 db; a.f.c.



Racal Type RA9 single-sideband receiver for ground-to-air communications in the band 2.5 to 20 Mc/s.

capture range (deviation from nominal transmitter frequency over which control is effective) 600 c/s with filter in and 3 kc/s without; a.f. output 2 watts (for 2.5% distortion) with 3 or 15 ohms load and 1 mW into 600 ohms load for P.O. lines. A self-contained monitor loudspeaker is operated at about 0.25 W.

The makers are Racal Engineering, Ltd., Western Road, Bracknell, Berks

F.M. Tuner Coils

A SET of six coils for the f.m. tuner described by S. W. Amos and G. G. Johnstone in our April and May issues this year is now obtainable from Copy Windings, Healey Lane, Batley, Yorkshire, the price being 47s 6d. They are wound on the specified sizes and types of former with all windings correctly dimensioned and securely fixed in position.

The ratio detector transformer, together with two GEX34 crystal diodes and capacitors C_{2a} , C_{30} , C_{31} and C_{32} are neatly packed into a $\frac{13}{16}$ -in square can measuring $2\frac{3}{8}$ in high. The two i.f. transformers are assembled in the same size of can and the capacitors C_{24} , C_{21} , C_{24} and



Set of coils for WIRELESS WORLD F.M. Tuner made by Copy Windings.

 $C_{_{2\,\delta}}$ are actually of closer tolerance than specified, being $\pm\,2\,\%,$ which is all to the good.

Oscillator, r.f. and aerial coils are each housed in $\frac{13}{15}$ -in square cans $1\frac{3}{5}$ in high, and in the samples sent to us C_1 was included with L_1 and L_2 , but C_9 was omitted from the oscillator assembly.

Electrostatic " Tweeters "

DESIGNED for frequencies from 5 to 20 kc/s, the LSH75 loudspeaker unit is of the two-electrode type with a gold-sputtered polythene diaphragm and perforated metal fixed plate. It measures $3in \times 3in \times 4in$ and weighs 14 ounces. Normally it is connected to the anode of the output stage through a simple high-pass R-C filter and is polarized from the h.t. supply (max. working voltage 300). The capacitance is 800pF and the power-handling capacity in the designed frequency range is up to 6 watts.

the designed frequency range is up to 6 watts. For powers up to 10 watts, the LSH100 is recommended by the makers. This measures $5in \times 4in \times \frac{3}{4}in$ and is enclosed in a composite metal and plastic case with louvres for the sound outlet; the weight is $3\frac{1}{2}$ ounces.

for the sound outlet; the weight is $3\frac{1}{2}$ ounces. The price of the LSH75 is 12s 6d and of the LSH100, 21s. The units are handled by Technical Supplies Ltd., 63 Goldhawk Road, London, W.12.

Type LSH75 and (right) LSH100 electrostatic loudspeakers.



MICROWAVE HARBOUR BEACON

Navigational Aid Using Transistors for Small Craft

By A. L. P. MILWRIGHT*

HE beacon described here was developed for the Ministry of Transport and Civil Aviation by the Admiralty Signal and Radar Establishment as an aid to entering harbour in bad visibility for craft which are too small to carry a standard marine radar, such as are found in the herring drifter fleet. The system is an adaptation of the old Lorenz type of landing aid for aircraft and consists of a 3-cm radar type transmitter mounted at a harbour entrance and radiating from two aerials which have overlapping beams. The transmitter is so sited that the line of intersection of the two beams is along the safe course line for entering harbour, as shown in Fig. 1. The transmissions consist of pulses of approximately 0.25 µsec duration with a repetition rate of 1,000 p.p.s. The output of the transmitter is switched in turn to each of the aerials in such a sequence that the morse letter B $(-\cdot \cdot \cdot)$ is transmitted from one aerial and the morse letter V (\ldots) from the other aerial. The characters of one letter are transmitted during the period of the space intervals between the other letter with the result that along the line of intersection of the two beams where the amplitude of the signals from each aerial is equal, a continuous signal is received as is shown in Fig. 2.

The shore-based transmitter and modulator unit are mounted immediately behind the aerial (Fig. 3) and consist of a free running blocking oscillator (CV73) which produces a 0.3-µsec pulse at 1,000 p.p.s. This is cathode coupled to a hard valve modulator (CV73) the output of which is capacity coupled to a magnetron (M503). The magnetron delivers a pulse of 0.25 μ sec duration with a peak power of 7 kW and the output is fed to each of two aerials in turn via a waveguide switch. This switch consists of an H-plane four-arm cross junction with a vane mounted across the junction at 45° . In order to switch the power to each aerial in turn, the vane is pivoted about the centre of the junction through 90° by a solenoid and lever arm, and the solenoid is energized via a pair of contacts and a cam driven by a small synchronous motor. The cam is cut to operate the solenoid in the time sequence ... - The fourth arm of the switch is terminated in a short circuit to absorb any leakage past the vane into the other aerial. A monitor of the power output is provided by a meter and thermistor bridge which is coupled to the magnetron output by a waveguide directional coupler.

The aerial consists of two vertical 2ft 6in linear resonant slot arrays mounted at one end and on each

^{*} Admiralty Signal and Radar Establishment.





Fig. 1. Approximate coverage of the microwave harbour beacon showing the narrow equi-signal course path.

SIGNAL FROM "B"AERIAL	
SIGNAL FROM "V"AERIAL	
RECEIVED SIGNAL IN EQUISIGNAL ZONE	

Fig. 2. The equi-signal course path arises from the overlapping of the B and V morse signals radiated by the two aerials.



Fig. 3. Transmitter and aerial system (only one aerial visible) of the microwave beacon.

side of a separator plate measuring $4ft \times 3ft$. The horizontal radiation diagram consists of two lobes each 60° wide overlapping over a narrow sector.

The system involves an amplitude comparison between the transmissions from each half of the aerial and since the ear cannot detect changes in amplitude of less than 1db the effective angular width of the equi-signal sector is 1°. The vertical beam width is 2.6° .

The ship's receiver is a wide-band pre-tuned crystal receiver coupled to a small horn aerial having hori-



Fig. 4. The ship's receiver showing (a) the complete set and (b) chassis (with horn) removed from the case.



zontal and vertical beam widths of 25° . The receiver and its power supplies are built around the horn and measure $4in \times 5in \times 5in$ (Fig. 4a and b) and weigh $4\frac{1}{2}lb$. The receiver's amplifier has a gain of 84db and uses four junction type transistors. The overall sensitivity is 10^{-8} watts, and reception is by means of headphones. The power supplies are provided by a $4\frac{1}{2}$ -volt flashlamp battery and the current drain is 5 mA. An improved model of the receiver (Fig. 5) incorporates a horn having a horizontal beamwidth of 16° and it is supported on a rotatable searchlight mounting. This is intended for fitting through the wheelhouse roof of a small boat. The output of this receiver is brought out to a 'phone jack on the rotating arm into which a loudspeaker attachment may be plugged. The loudspeaker attachment may be

Trials and demonstrations have been carried out at Frazerburgh and Arbroath and the equipment operated very satisfactorily out to a range of approximately seven miles.

[A commercial version of the microwave beacon described is, we understand, being made by Elliott Bros., Ltd.—ED.]

Record and Stylus Wear

INCREASING the weight on a stylus point in contact with a record reduces the pressure. This apparent paradox is resolved when it is realized that plastic yielding of the record material increases the area of contact at a greater rate than the increase of force causing the deformation.

This point is made early in a paper by F. V. Hunt in the Journal of the Audio Engineering Society (Vol. 3, No. 1, January 1955) "On Stylus Wear and Surface Noise in Phonograph Playback Systems." Prof. Hunt makes a detailed mathematical analysis of the stress contours under a spherical indenter in contact with the plane surface of a plastic material and sets out the conditions for the onset of plastic yielding. This is influenced by the size of the point to the extent that a perfect specimen of a material, without flaws such as crystal lattice imperfections, has considerably greater strength, and the probability of finding such a specimen under the contact increases as its size is reduced. This effect is observed in the results of "micro-hardness" indentation testing of materials, and may account for the fact that microgroove records are not excessively damaged in spite of the fact that the 2:1 reduction in recommended playing weight reduction as compared with standard records is only half what is should be (the stylus *force* must vary directly as the square of the stylus radius to maintain constant stylus pressure).

The theory is extended to the case of real surfaces making contact only at a few high points or asperities, and it is shown that on this scale plastic yield is intermittent and accounts for much of the surface noise. A discontinuity in the load/wear curve of a stylus exists at low loading and spectacular increases in the life of records and styli can be expected if, for example, the playing weight on a point of 0.001 inch radius is reduced below 2 grams.

The rupturing of bonds of adhesion at points of contact between local asperities on the stylus and record groove surface is thought to account for the present irreducible minimum of surface noise. These bonds are often as strong as the cohesive forces of the materials themselves and the shear stresses required to break them combine to give the coefficient of friction of the material.

Choice of materials having low inter-facial adhesion (and a low coefficient of friction) should, on this argument, have very low surface noise, and the suggestion is made that "Teflon" (polytetrafluoroethylene) might be tried as a record material.

Institution of Electrical Engineers

Institution of Electrical Engineers Radio and Telecommunication Sec-tion.—November 3rd. "The new high-frequency transmitting station at Rugby" by C. F. Booth and B. N. MacLarty. November 9th. "A transistor digital fast multiplier with magneto-strictive storage" by Dr. G. B. B. Chaplin, R. E. Hayes and A. R. Owens. November 15th. "An electrolytic-tank equipment for the determination of electron trajectories, potential and gradient" and "A method of tracing electron trajectories in crossed electric gradient" and "A method of tracing electron trajectories in crossed electric and magnetic fields" by Dr. D. L. Holl-way to be read by Dr. J. H. Westcott. November 21st. Discussion on "The reception of Band I and Band III television programmes" opened by E. B. Westbau

E. P. Wethey. November 29th. "The specification of the properties of the thermistor as a circuit element in very-low-frequency systems" and "A vector method for amplitude-modulated signals" by Dr. C. J. N. Candy.

All the above meetings will be held at Savoy Place, Lordon, W.C.2, at 5.30. Cambridge Radio and Telecommuni-

cation Group.—November 8th. Ad-dress by H. Stanesby (Radio Section chairman) at 8.15 at the Cavendish

chairman) at 8.15 at the Cavendish Laboratory, Canibridge. North-Eastern Centre. — November 14th. "A Transatlantic Telephone Cable" by Dr. M. J. Kelly, Sir Gordon Radley, G. W. Gilman and R. J. Halsey at 6.15 at the Neville Hall, Newcastle-on Tune.

on-Tyne. North Midland Centre.—November 1st. "A Transatlantic Telephone Cable" by Dr. M. J. Kelly, Sir Gordon Radley, G. W. Gilman and R. J. Halsey at 6.30 at the Central Electricity Authority, 1, Whitehall Road, Leeds. North-Western Radio and Telecom-

interventional and relecom-munication Group.—November 9th. "High-speed electronic-analogue comput-ing techniques" by Dr. D. M. MacKay at 6.45 at the Engineers' Club, Albert Square, Manchester.

at 6.45 at the Engineers' Club, Albert Square, Manchester. North Scolland Sub-Centre.—Novem-ber 9th. "Thermionic valves of im-proved quality for Government and industrial purposes" by E. G. Rowe, P. Welch and W. W. Wright at 7.0 in the Electrical Engineering Dept., Queen's College, Dundee. South Midland Radio and Telecom-munication Group.—November 28th. "A radio position fixing system for ships and aircraft" by C. Powell at 6.0 at the James Watt Memorial Institute, Great Charles Street, Birmingham. Reading District.—November 28th. "The physics of transistors" by Dr. E Billig at 7.15 at the George V Room, George Hotel, King Street, Reading. Tees-side Sub-Centre. — November 2nd. "The Manchester-Kirk o'Shotts television radio relay system" by G. Dawson, L. L. Hall, K. G. Hodgson, R. A. Meers and J. H. H. Merriman at 6.30 at the Cleveland Scientific and Technical Institute, Midelesbrough. Graduates' and Students' Section.— November 2nd. "Communication of in-formation—human, animal and machine" by Dr. E. Colin Cherry at 6.30 at Savoy

November 2nd. "Communication of in-formation-human, animal and machine" by Dr. E. Colin Cherry at 6.30 at Savoy Place, London, W.C.2. November 15th. "An introduction to the transistor" by A. V. Bryant at 7.0 at R.E.M.E. Training Centre, Bailleul Barracks, Arborfield, Berks. November 29th. "Colour television" by L. A. Harris at 7.0 at the Public Library, Chelmsford.

WIRELESS WORLD, NOVEMBER 1955

MEETINGS

British Institution of Radio Engineers

London Section.—November 30th. "High-fidelity loudspeakers—the per-formance of moving-coil and electro-static transducers" by H. J. Leak at 6.30 at the London School of Hygiene and Tropical Medicine, Keppel Street, W.C.1. Merseyside Section.—November 9th.

"The Decatron tube in a digital trans-mission system" by G. Shand at 7.0 at the Chamber of Commerce, Old Hall Street, Liverpool, 3. North-Eastern Section.—November

9th. "Turret tuners for multi-channel television reception" by R. Holland at 6.0 at Neville Hall, Westgate Road,

television reception" by R. Holland at 6.0 at Neville Hall, Westgate Road, Newcastle-upon-Tyne. North-Western Section.—November 3rd. "Ground controlled approach and instrument landing system" by R. H. James and N. MacKinnon at 7.0 at the College of Technology, Sackville Street, Manchester.

Scottish Section.-November 10th "Some applications of electronics to marine echo-sounding" by F. Baillie at 7.0 at the Institution of Engineers and Shipbuilders, 39 Elmbank Crescent, Glasgow, C.2.

British Sound Recording Association London.—November 18th. "The de-velopment of professional recording equipment" by W. S. Barrell at 7.0 at E.M.I. Studios, 3 Abbey Road, N.W.8.

Television Society

London.-November 11th. "The application of semi-conductor diodes to television circuits" by J. I. Missen. November 24th. "Interference with

November 24th. "Interference with television reception: its causes and cures" by R. A. Dilworth. Both meetings are to be held at 7.0 at 164, Shaftesbury Avenue, W.C.2.

Physical Society

Acoustics Group.—November 11th. "Acoustic interferometers" by Dr. E. G. Richardson at 5.30 at Imperial College. South Kensington, London, S.W.7.

British Kinematograph Society

London.—November 23rd. "The display of television pictures in the home" by E. P. Wethey at 7.15 at the Gaumont-British Theatre, Film House, Wardour Street, W.1.

Radio Society of Great Britain London.—November 11th. "Com-pressed beams" by G. A. Bird (G4ZU) at 6.30 at the I.E.E., Savoy Place, W.C.2.

DO YOU KNOW?

HOW long the elements for a Band II or Band III aerial should be? Where to apply for a licence to estab-

lish a mobile radio-telephone link? What value of resistors to use for a television attenuator?

The address of each of the above

societies? The base connections for a PCF80?

The answers to these and very many other radio technical and organizational ouner radio technical and organizational queries can be readily given by the owner of a Wireless World Diary. The 1956 edition—the 38th—is now obtain-able from booksellers and newsgents, price 5s 10d (morocco leather) or 4s 1d (rexine).

In addition to the 80-page reference section, which includes details of base connections for 600 valves, there are the usual week-at-an-opening diary pages.



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RANDOM RADIATIONS

Band III Aerials

INGENIOUSLY designed Band III and two-band aerials continue to appear and I expect there'll be lots more new ideas in months to come. Amongst the earliest to make their bow were the J-Beam single-slots and slot-beams for Band III. As "ghosting" is one of the big problems in that band the slot, either as a "single" or with reflector and directors, should be a help to good reception in difficult places. All kinds of clip-on attachments are being produced for converting a Band I aerial for two-band operation. Many will be useful only if the B.B.C. and I.T.A. transmitters lie in the same direction from the viewer. But there are also several clever ideas for enabling the two arrays to be oriented independently. One that interests me is the Antiference "electronic coupling." An arm carrying the desired number of Band III elements and the Band I dipole, critically spaced from the nearest of these, is mounted on the mast and turned in the direction of the I.T.A. station. A second arm carries the Band I reflector and can be swung so that it and the dipole are directed towards the B.B.C. station. Both arrays are thus properly oriented and only a single feeder, in which both signals appear, is needed.

Exhibition Prefabs?

THE bright suggestion has been made that stands for future wireless shows should be prefabricated. The idea is that each exhibitor would make his own stand in sections. Every section would be completely wired and the stand would be erected by the simple process of putting the sections in their right places and plugging each section into the next. All the wiring would terminate in a single heavy-duty plug and only one main electrical connection would then be needed for any stand. It seems quite a brilliant idea; but one wonders what kind of reactions it would provoke!

Lighting or Power?

A STORY came my way the other day which was declared to be true by its teller. It might well be, for there's no end to man's (and

woman's) ignorance and folly about things electric. Anyhow, here it is. Into an electrical dealer's shop walked a man who had just moved into a house in the neighbourhood. "There are some wiring changes and additions that I want made," he said, "and I'd be glad if you'd take on the job." The dealer having replied suitably, his visitor went on : "Oh, by the way, could you lend me an ammeter for about half an hour?" Asked what he wanted it for, he replied that there were a good many sockets in the house and he wasn't quite sure which of them were for lighting and which for power. "But how will the ammeter help you?" "Oh, don't you know? It's quite simple. You just put the prods into the socket: if it's for lighting the meter reads 5 amps; if for power it reads 15.3

Needlessly Scrapped

ONE of the things I'd like to know (though it's not likely that I or anyone else ever will) is the number of television c.r. tubes perfectly capable of giving good service that are scrapped in a given year. It would most certainly run to a largish figure. The two main causes of unnecessary replacement are cathode-to-heater

shorts and low emission. Nothing can be done about the first of these if the mains supply is d.c. but if it is a.c., fitting an inexpensive isolating transformer for the heater puts things right. The owner of a c.r. tube with low emission has to take a chance if he's on d.c.; reactivation by running the cathode at high temperature for brief periods may be successful and give the tube a further lease of life, but it may also mean a burnt-out heater. It's a chance worth taking anyhow, for an otherwise useless tube may last for months if the reactivation is successful. The a.c. man can usually count on good results. All that he needs is a booster transformer for the heater. I know one c.r. tube which still gives an excellent picture 18 months after the fitting of a booster.

Steps Forward

AMONGST the improvements one has so far noticed this year in TV receivers is the increased adoption This should of flywheel sync. lighten the lot of viewers who live in places where interference of certain kinds is bad and those whose homes are in fringe areas with a poor signal-to-noise ratio. Another welcome improvement is the wider use

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of effective automatic gain control. It seems to me, too, that a good many new sets are making a better job of interlacing than did some of their predecessors. But I must say I'd like to see more rather than less genuine a.c.-only television sets with double-wound mains transformers in their power-packs. Diehard or not, I'll never become reconciled to the transformerless receiver, worked as it is more often than not from a twopoint a.c. mains socket.

CLUB NEWS

Barnsley.—At the meeting of the Barnsley and District Amateur Radio Club on November 11th, W. Williams will speak on "Band II f.m. reception." A fortnight later F. Robinson (G3FLQ) will describe a superhet double conversion adaptor. Meetings are held at 7.0 at the King George Hotel, Peel Street, Barnsley. Sec.: P. Carbutt (G2AFV), 33 Woodstock Road, Barnsley, Yorks.

Birmingham.—The annual dinner of the Midland Amateur Radio Society will be held at the Imperial Hotel, Birmingham, on November 12th. Sec. C. J. Haycock, 360 Portland Road, Edgbaston, Birmingham, 17.

Birmingham, 17. Cambridge.—The Cambridge University Wireless Society (G6UW) meets at 8.15 each Tuesday of the full term in the Cavendish Laboratory. Sec.: A. Brunnschweiler, Pembroke College, Cambridge.

Cleckheaton.—During November the Spen Valley and District Radio and Television Society will be visiting the psychology department and seeing the electron microscope at Leeds University (2nd) and the medical physics department of the General Infirmary, Leeds (30th). On the 29th Mullard films will be shown at the Metropole Hotel, Leeds. Meetings commence at 7.30. Sec.: N. Pride, 100 Raikes Lane, Birstall, Nr. Leeds.

stall, Nr. Leeds. Edinburgh.—The first of a series of talks describing "A Beginner's Transmitter" will be given on November 3rd to members of the Lothians Radio Society. On the 17th J. W. Kyle (GM6WL) will deal with 70-cm reception. Meetings are held at 7.30 at 25 Charlotte Square, Edinburgh, 2. Sec.: J. Good (GM3EWL), 24 Mansionhouse Road, Edinburgh, 9. Southend—At the meeting of the

Road, Edinburgh, 9. Southend.—At the meeting of the Southend and District Radio Society on November 11th, H. Wilkinson will talk on "Current trends and developments in radio and television." On November 25th a representative of the Automatic Coil Winder Company will speak on "Recent developments in test equipment for the electronic industry." Meetings are held at 7.15 at the Ekco Works, Southend-on-Sea. Sec.: P. C. Baldwin, 13 Inverness Avenue, Westcliff-on-Sea, Essex.

Swindon.—Regular meetings are now being arranged for the recently formed Swindon Amateur Radio Club, of which R. Reynolds (G3IDW) is chairman. Weekly classes of instruction in preparation for the amateur radio examination are being held at the College, Swindon, at which the instructor is G. R. Pearce (G3AYL), who is secretary of the club. His address is 102 Kingshill Road, Swindon, Wilts.



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UNBIASED

Electronic Tapostat

IN the August issue of *Wireless World* the Editor demanded the invention of an electronic "bookmark" or tapostat for tape recorders so that a user could rapidly wind-on or rewind the tape until he found the particular section he desired to play.

Naturally, I replied by return of post detailing a scheme. This provided for the tape to have supersonic frequencies imposed on it at selected spots during recording. The great increase in tape speed on fast wind or rewind would naturally convert such supersonic frequencies into radio frequencies. These could be picked off the tape by a tuned r.f. head, amplified, rectified and used to actuate a relay-operated brake which would stop the mechanism at any desired spot on the tape.

The Editor rejected my idea on the ground that it was several years in advance of its time. Those were not his exact words but they were certainly implied in his answer. He told me that our sound-recording tapes have too noisy a background for my idea to succeed. It cannot be doubted, I think, that in time all tape recordings will have the same silent background as the tapes being used in the experiments for the recording of television and then, of course, my electronic tapostat will come into its own.

"Walter, Lead Me to the Altar"

READERS of *Wireless World* will have heard of Dr. Grey Walter of encephalograph fame and be familiar with his important work in the field of neurological diagnosis and kindred subjects.

According to some sections of the daily press, however, Dr. Grey Walter has now started to trespass on the prerogatives of Madame Estelle by warning us of matrimonial pitfalls ahead. To give him his due, however, his matrimonial soothsayings, unlike those of Madame Estelle, are founded on the solid rock of science. He does not tell girls to beware of the buccolingual blandishments of sinister dark-haired men of the "Sir Jasper" type. Instead he advises them to demand that any would-be Romeo should produce his encephalogram.

Dr. Grey Walter advises that only those with compatible encephalograms should launch out together on the stormy seas of matrimony. It would seem advisable for all budding benedicks of either sex to carry around with them, like a passport,

By FREE GRID

their personal encephalograms, duly authenticated by Dr. Grey Walter and his colleagues. Thus girls of the future will have as their theme song the same ditty which was so popular among their Edwardian sisters but with a slightly different meaning; "Walter, Lead me to the Altar."

Bi-Fi

IN my opinion the outstanding audio feature of 1955 has been the introduction by more than one firm of



East Anglian 2-channel broadcasting

stereophonic reproduction from records. The name "stereophonic," or any of its obvious variations, is rather a mouthful, however, and I think, following the tendency of modern jungle jargon, a better name would be bi-fi. The whole raison d'être of the thing is that it is not only binaural—for all listening is that—but "bi" everything else as two channels must be used throughout. It is also bi-fi in the hellenic sense of "bi," for the music and everything else does indeed appear to be living.

In view of the great success of recent bi-fi demonstrations, I hope that the director of technical services at the B.B.C. now takes a more enlightened view of bi-channel broadcasting than he did in his letter to *Wireless World* a few years ago (April, 1950). When the Norwich v.h.f. station opens in 1956 he will have a unique opportunity of experimenting with bi-fi.

According to the B.B.C.'s published map, those living between Ipswich and Colchester and in a corridor extending west beyond Cambridge will be able to receive both the Wrotham and Norwich transmissions. To make matters quite clear I reproduce part of the v.h.f. map on which I have shadedin the corridor. One or more of the three transmitters at these two stations will always be radiating the same programme and it will, therefore, be possible to experiment with bi-fi in that area at a minimum of expense as it would only be necessary for the B.B.C. to use separate mikes and associated gear to feed the two chosen transmitters. It could at least be tried out with the two transmitters radiating the Third Programme where bi-fi would be most appreciated. What about it, Sir Harold?

Strabismogenic TV

THE size of the screen in domestic television receivers has been growing steadily during the past few years to such an extent that many of the latest sets seem only suitable for baronial halls. Unfortunately the feeling that they must "keep up with the Joneses" forces people into the same foolish competition of "bigness" in which they once indulged in sound radio, when it seemed everybody's wish to compete with the bulls of Bashan.

To attempt to view some of these huge screens in the confines of an ordinary room leads only to strabismus, as I found out the other day when invited to a friend's house to view his latest "horror-scope." Eventually people will learn common sense as they did with sound radio.

Now even with a 17-in screen, lines are painfully evident. One or two firms do provide spot wobble in this and larger sizes but it is high time something was done about this problem, by the radio industry in general, without resorting to the drastic remedy of altering the British standard of 405 lines.

The disadvantages and complications of this latter remedy have been



405 hard lines

pointed out more than once in this journal, but surely it is not beyond the wit of our technological tycoons to tackle this problem with the same vigour and success as women remove the lines from their faces.

he AVO Valve Characteristic Meter Mark III offers the Radio Engineer far more than is generally implied by the words "a valve tester".

This compact and most comprehensive Meter sets a new high standard for instruments of its type. It will quickly test any standard receiving or small transmitting valve on any of its normal characteristics under conditions corresponding to a wide range of D.C. electrode voltages.

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2

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3

E180F wide-band low-noise



PRINCIPAL CHARACTERISTICS

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Vg2	• • •		160V	ra	 	35k Ω	
Vgl			- I V	μ _{gl-g2}	 	50	
la	• •		I3mA	Heater	 	6.3V,0.3A	
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Input damping = $6k\Omega$ (at f = 50 Mc/s and with cathode connections strapped).





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Type		Length (mm)	Dia. (mm)	Volts	Amps. (Approx.)	Max. Pea Inverse Vol (Volts)	Max. Pea Forward Vo (Volts)	Max. I Current	Max. 1 Current	Approx. Tube Drop (Volts)	Peak Power Level (a)	American Equiv.	C.V. Nu
AFX.212	Xe	54	19	6.3	0.25	350	350	0.11	0.025	11	-	6D4	1949
AFX.203	Xe	176	57	2.5	4.0	300	280	I.7	0.40	II		ČіА	2868
FX.215	\mathbf{H}_2	286	97	2.5	27.5	16,000	16,000	200	0.20	100(b)	2.0 x 10 ⁹	—	2203
FX.219	\mathbf{H}_{2}	222	65	6.3	10.6	16,000	16,000	325	0.20	100(b)	3.2 x 10 ⁹	5C22	2520
FX.225	\mathbf{H}_{2}	175	65	6.3	6.1	8,000	8,000	90	0.10	100(b)	2.0 x 10 ⁹	4C35	1787
FX.227	H ₂	132	40	6.3	2.25	3,000	3,000	35	0.045	100 (b)	0.3 X 10 ⁹	3C45	372
• • • •				,	Ke — Xen	071		H2 HV	drogen				

(a) — Product of peak forward voltage, peak current and pulse repetition frequency.
 (b) — Under conditions of pulse operation.

ENGLISH ELECTRIC VALVE CO. LTD.

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The Automatic Frequency Monitor (20 Mc/s) is but one of a series of high grade monitors now in course of manufacture for the accurate measurement of frequency.

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The result, in decimal notation, is presented on eight panel mounted meters each scaled from 0 to 9 and the unknown frequency is automatically remeasured every few seconds.

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We are proud of the vast number of our loudspeakers incorporated in radio and television receivers used throughout the world. Their quality of reproduction and unfailing performance have been amply proved over many years in every climate and condition of service.

Rola Celestion Ltd

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First things last!

We freely confess that when we began planning the Tannoy high-fidelity Domestic Sound System, we did not contemplate designing and manufacturing our own phono-cartridge.
After all, there were several excellent cartridges already on the market. We changed our minds when the first power amplifier and 'Autograph' pre-amplifier began to emerge from the assembly lines. It was evident that the almost uncanny faithfulness of the amplifying chain, ending with the now world-famous Dual Concentric Speaker, justified the use

of a cartridge of more than ordinary 'freedom'. Our engineers therefore set

about the task of designing a cartridge which was free to extract, and pass on, *everything* from the recorded groove without any spurious effects or blanketing resonances. The result is the Tannoy Variluctance Cartridge ... a precision transducer which would be completely wasted on any but the very finest of high-fidelity reproducing equipment.



Practitioners in Sound



SPECIFICATION

Every cartridge hand-made and laboratory tested Frequency response within 2 dB to 16,000 K.C. No resonant peaks No undamped resonances in sub-supersonic range Simple turn-over mechanism Stylus assemblies completely independent Instantaneous replacement of styli without use of tools Optimum lateral to vertical compliance ratio Very low effective dynamic mass Output : 20 mV at 12 cm. per second Termination load : 50,000 ohms. Tracking weight : 6 grams for all discs Available with either diamond or sapphire styli

VARILUCTANCE Phono-Gartridge

CARDINAL PROPERTY IN THE REAL PROPERTY. They say we make a perfect pair...

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Contraction (1990)

To be exact, this is the 12 pin version of the Multi-Way Plug and Socket range, which covers 4, 8, 12, 20 and 28 ways. The range features unusually low insertion pressures, and embodies considerable experience in meeting humid conditions. Designed to overcome as far as possible the difficulties encountered when using this type of connector in rack mounting applications, they have greater latitude in matching up than any comparable product, and are in use throughout the world in Radio, Television and Telecommunications equipment by such renowned firms as :---Messrs. Marconi's Wireless Telegraph Co. Ltd., The English Electric Co. Ltd. and Messrs. Standard Telephones & Cables Ltd.

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PHONE: NEW 3181/2/3

P.C. 2



Are you the auto-excited type, old boy?

Very, George. I remember a straight-eight I used to have, with four synchronised carburettors and temperamental brakes, that once . . .

No, no--what I mean is, are your rectifiers connected in series with the output winding so that the output current automatically saturates the transductor \ldots .

Where do you get all this 'output' on transductors, George?

From a Parmeko leaflet—it tells me that they have a standard range of 14 types for operation at supply frequencies of 400 and 1600 cycles and some for 50 cps.

Sounds very varied.

They are indeed-they're being used in magnetic amplifiers all over the place.

Do you think I could get one of these leaflets of theirs?

Yours for the asking, George. They're in Two Colours and Fully Illustrated, as they say . . .

PARMEKO of LEICESTER

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Model BS

by

NOVEMBER, 1955



new world-beating SCOTCHBOY EXTRA BOY Magnetic recording tape 190m

Tough, thin, polyester base

GIVES UP TO 6 HOURS PLAYING TIME

- packs 1,800 feet on normal 1,200-foot reel.

THE FINEST BASE-FILM EVER MADE

The astonishing new polyester base-film for 'Scotch Boy 190M,' is so much stronger than other tape bases that it can be made $33\frac{1}{3}\%$ thinner—and still be stronger. This means you get 50% more length—and 50% EXTRA PLAYING TIME—on the same-sized reel.

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THE WORLD'S FINEST TAPE

'Scotch Boy 190M' has been developed and produced in Britain by the 3M Company. Its appearance in Britain is its first appearance in the world. This is a landmark in the development of tape recording.

SCOTCH BOY' 190m

MAGNETIC RECORDING TAPE with polyester base

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NOVEMBER, 1955

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NOVEMBER, 1955



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NOVEMBER, 1955



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P420	5.5 µµf, 10 megohms	1 v/cm			
P450-L	2.5 µµf, 10 megohms	2.5 v/cm			
P4100	2.5 µµf, 10 megohms	5 v/cm			



Type 545 Oscilloscope Characteristics

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VOLTMETER

STABLE · ACCURATE · TEN RANGES

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Specification

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Ten-way switch giving full-scale ranges of 0.01, 0.03, 0.1,
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For wide angle tubes with 38 mm. diameter necks.

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1C5GT	3Q4	6BG6G	6SA7 6SA7GT	8D2	23D	84	826	4260A	CV57	DRM2B DRM3B	GDT4B
1D5	3458	6BH6 6BJ6	6SA70T	8D5 9D2	OF A RICL	88J 89(¥)	828 929A	42020	CV 30	DRMOB	GEX00 GEX34
1D6 1D8/GT	ONOUT.	6BR7	45070T	9D6	24G 25A6/G 25A6GT	10071	929B	4205E 4212E 4260A 4313C 4328D 4378	CV84 CV67	E1149	GEY35
100/01	3Q4 3Q5G 3Q5GT 3S4 3V4	6BS7	6SC5 6SC7GT 6SD7GT	9HP7	25L6 25L6GT 25SN7GT	100TH 117L7GT 117N7GT 117N7GT 117Z6GT 210HL 210SPG	8308	4690	CV72 CV75	E444S E1148 E1155	GEX35 GEX44/1 GEX45/1
1E7G 1E7GT	4C27	6BW6	6SF5	10	25L6GT	117N7GT	830B 832	4690 5763	CV75	E1190	GEX45/1
11760	4C29	6BW7	6SF7	10 7	25SN7GT	117Z6GT	832A	7193	CV83	E1191	GEX54
1040T 1050	4D1	6BX6	68G7	10D1	25¥5 25Z4G	210HL	833/833A	7475	CV85 CV88	E1191 E1192	GEX54 GEX54/3
1 G 5G	4353	6C4	6SH7	11D3	25Z4G	210SPG	836	8011	CV88	E1231 E1248	GEX54/4
106/ GT	4THA	6C5	68H7GT	11D5	25Z5 25Z6G	210SPT 210VPT	837	8012A 8013A	CV92 CV100	E1248	GEX54/5 GEX55/1
1H5G	4TPB	6C5G. 6C5GT	6SJ7	12A6	25Z6G	210VPT	838	8013A	CV100	E1254	GEX55/1
1H5GT	5AP1	8C5GT	68J7GT 6SJ7Y	12A6GT 12A8GT	25Z6GT 27	212E	841 843	8016	CV101 CV118	E1265 E1266	GEX64 GEX66 GEX69
1H6G	5A/102D	606	054/1	12AH7GT	28D7	215P 2155G	850	8019 8020	00110	E1200	CEVED
1L4 1LA6	584G 58/502A 58P1 5CP1	6C21 6CD6G	65K7 65K7GT 65L7GT 65N7GT	124 18	30	21330	860	9001	CV 125	1978	GLARRA
1LC6	KPD1	6CH6	AST.7GT	12AH8 12AT6	32	217C 220B	881	9002	CV172	E1278 E1320	GL466A GL451 GT1C GU20
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1N5G	5D21 -	6 E 6	6997	12AX7	35L6GT	231D 250TH	866A 866JR	AC4/PEN	CV415 CV967	E1379	GU50 H30
1N5GT	5GP1	6F5 6F5G	6ST7 6T7G 6U5G 6U5/6G5 6U7G	12BA6	35T	250 TH	866JR		CV967	E1438	H30
1P5GT	5GP1	6F5G	6T7G	12BE8 12BH7	35TG	262A/B 279A 282A	869B	ACT6 ACT17 APP4B APP4C APP4C AR12	CV980	E1468	H63
105GT 184	5JP4 5L35 5LP1 5R4G Y	8F5GT	6U5G	12887	35704 3523 35240T 85250T	279A	872A 874	AUT17	CV988 CV1481	E1474 E1481	HD14 HF30
1R4	5135	6F6 6F6G 6F6GT	605/665	1208	3343	304TH. TL	875A	APP4D	CV1593	E1481	HF30
1R5 184	OLP1 FR4CW	0010	0070	12C8GT 12DP7	2575GT	307A	876	APPAG	CV1588	E1494 E1496	HL2 HL2K HL4
184	5T4	6F7	AVAG	1286	38	310A	878A	AR12	CV1598	EA50	HL4
174	5U4G	6F7E	6V6 6V6G 6V6GT	12J5GT 12J7GT	87	310A 310B 311A 313C -323A	884	AR13 AR300A	CV1588 CV1598 CV6008	EB34 EB91	HL23
105	5¥4G	6F8G 6F8GT	6W2 6W7G	12J7GT	38	311A	905A	AR300A	CY31	EB91	HL41
17	5X4G	6F8GT	6W7G	12K7GT	39/44	313C	923	AR4101	CY 32	EBC3	HP210
2A3	5Y3G 5Y3GT	8G5G		12K8	40	-323A	931A	ARP3 ARP4	D1	EBC33	HR210
2A4G	5Y3GT	6666	6X5 6X50 6X50 6X50 6Y60	12K8GT 1207GT 1207GT 125A7 125A7GT 125C7 125C7 125G7	41	327A	954	AKP4	D15 D41	EBC41	KMV6
2A5	5¥4G	6H6 6H6G 6H6GT	6X50	120701	41MP 41MPT 41MTL 41MXP 41STH	328A/4328A 337A	955 956	ARP13 ARP38 ARS6	D41 D42	EC54 ECC81 ECC82	KR8 KR6/3
2A6 2A7	5Z3 5Z4	OELOUT	evec	1984 707	AIMTT.	OFAT	957	ARSA	D43	FCC82	
2B7	5740	615	6770	12807	AIMXP	357 A	958A	AT4	1083	ECC83 ECC91 ECH22	KT2 KT8 KT24
2026	5Z4G 5Z4GT	6J5G	6Z5	12507	418TH	368A	959	AT15	D77 DA30 DA60	ECC91	KT8
2C26 2C26A	6A3	6J5GT	7A2		4.9	380A	991	AT40	DA30	ECH22	KT24
2C34	6A6	6J6 6J7	7A4	12SJ7	42SPT	388A	1299A	AT4 AT15 AT40 ATP4 AT870	DA60		KT30
2040	6A7	6J7	7A5	12SJ7GT	48 45	394A	1616	ATS70	DA90	ECH42	KT31
2043	6A8G 6A8GT	6J7G 6J7GT	7A6	128A7 128K7GT	45SPEC	3577 368A 380A 389A 394A 450TL 703A	1616 1619 1622	AU5 AU7	DA90 DA100 DAF91 DDR25	ECH42 ECL80 EF22	KT32 KT33C
2D21 2E22	6ABGT	6J8G	7A7 7B6	1251.707	48	7054	1624	AZI	DDR25	EF36	KT44
2J21A	6AB8	SKAG	787	125L7GT 125N7GT	50C5 50CD6G 50L6GT 50Y6GT	705A 707A/B	1625	AZ31	DET5	EF37	KT61
2134	6AC7	6K6G 6K6GT	787E 78P7	19807	50CD6G		1625 1626	AZ41	DET9	EF37A	KT66
2138	6AF6G	6K7	7BP7	12SQ7GT	50L6GT	709A	1690	B21	DET12	EF39	KT71
2J39	6AG5	6K7G 6K7GT	7C4	12SR7	50 YOGT	713A	1635	B30	DET16	EF41	KTW61
2J39 2J48	6AG7	6K7GT	705	12SQ7GT 12SR7 12U5G	53A 53KU	714AY	1635 1642 1648	BL63 BT45	DET19	EF41 EF50 EF54	KTW61 KTW62 KTW63
2154	6AJ7	6K8	708	12X3 12Y4	5315.0	717A	1648	C5B	DET25	EF80	KTZ41
2J54B	6AK5	6K8 6K8G 6K8GT	7C6 7C7 7D5	1486	54 57	709A 709A 713A 714A¥ 717A 723A/B 724A 725A 726A	1851	010	DF91	EF91	KTZ63
2X2/879 2X2A	6AK6 6AL5	6L5G	703	14E7	58	725A	1851 1960	C9A	DF91 DF92 D_103	EF91 EF92	KTZ63 KTZ73
2A2A 3A4	6AM5	6L6	7D8	1487	59	726A	2050	C9A CAG25 CAV25 CK1005	D.103	EF93	L2
3AP1	6AM6	AL.AG	7D9	14K7	61P	800	2051 2151	CAV25	DH76 DH77	EF94	L30
- 3B7/1291	6AQ5	6L6GA	7E5	14R7	71A	801	2151	CK1005	DH77	EF95	L63
3B24	6AT6	6L7	7E6	1487	72 73	801A	3951 4003A	CL38	DH81	EL22	L77
3B26	6AU6	6L7G	7E7	15D2	73	803	4003A	CMG8	Dd101	EL32	L610 LD210
3B/151A	6AV6	6N7	717	15E	75 76	805	4019A 4019B	CMG22 CMG25	DH107	EL33 EL41	LD210 LD410
3BP1	6B4G	6N7G	707	15R	76	807	4018B	0/10/20	DK91	ED41	10.10



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• "Symphony " No. I F.M. TUNER

pounds.

We are pleased to announce that after extended research our new High Fidelity F.M. TUNER has been placed on the market and is available for immediate delivery from stock. It incorporates the latest type of permeability-tuned coil assembly of advanced design housed in die-cast protective anti-radiation design housed in die-cast protective anti-radiation shroud. The Tuner employs the most modern types of valves newly developed especially for F.M. circuits -ECC85, 2 x EF89, EB91.

The efficiency of the general circuit ensures extreme sensitivity and a very high music-noise ratio. The output impedance is $\frac{1}{2}$ Megohm, rendering it suitable for feeding into any normal amplifier including those of the Highest Fidelity class.

A volume control is incorporated to adjust for varia-tions in signal strength and amplifier input sensitivity. A radio/gram. selector switch and pickup socket are also incorporated, and the unit is readily linked to an A.M. Tuner witchout external chargeover switch. The slow-motion tuning drive is especially smooth and free from backlash and the glass dial is illuminated. Overall size is: 9in, wide x 6in, deep x 6in, high.

The power requirements are: 6.3 v. at 2 amps, and 250 v. at 40 mA. Our model FMI Power Pack is ideal for providing this power and has capacity for the average A.M. Tuner as well.

The price of this high grade F.M. Tuner is only 14 gns. tax paid, and the Power Pack £3/7/6 extra if required.

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MODEL TB as above, but with long pickup arm. Less heads, £8/11/-, post 2/6.

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"SYMPHONY" BASS RE-FLEX CABINET KITS. 30in. high, consist of fully-cut ‡in. thick, heavy, inert, non-resonant patent acoustic board, deflector plate, felt, acoustic board, deflector plate, felt, all screws, etc., and full instructions, Bin, speaker model, 85/-; 10in, speaker model, 97/6; 12in, speaker model, 45/7/6. The design is the final result of extensive research in our own laboratory and is your safeguard of optimum acoustic results. Carriage 7/6. Ready built 10/6 extra. results. Carria built, 10/6 extra.

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GOODMANS CORNER CAB INETS for the AXIOM ISO Mark 2 manufactured by us to Messrs. Goodmans' specification and ap-proved by Messrs. Goodmans. Height 44in. Price: complete kit Height 44in. Price: complete kit in plain board and lin. thick felt, 8 gns. Price: ready built, 10 gns. Finished in figured walnut, 26 gns. Other veneers to order. Carriage extra according to area. Ouoration by return.

Quotation by return. CONSOLE AMPLIFIER CABINETS. 33in. high, lift-up lid with piano hinge, take Tape Deck, Gram Unit or Auto-changer, Amplifier, Pre-Amplifier and Radio Feeder Unit finished medium walnut veneer. De Luxe version, price II gns. Oak or mahogany veneers 20 + extra. Special finishes to order. Carriage according to area, we will quote by return.

NOVEMBER, 1955

NEW ARCOLECTRIC SIGNAL LAMPS

For Low Voltage or Mains

Illustrated are a few signal lamps taken from our wide range. The insulation of every Arcolectric signal lamp will resist a flash test of 1,500 volts A.C. The S.L.90 illustrated here is a typical Arcolectric low voltage signal lampholder. It is designed to accept popular M.E.S. bulbs. The bulb is accessible from front or rear of panel. The domed plastic lens surrounded by a polished chrome bezel gives a most attractive panel appearance. This holder can be fixed in a single $\frac{3}{4}$ in. hole. The mains voltage signal lamp SL88/N is supplied complete with an M.E.S. neon tube and a suitable series resistance.

Write for Catalogue No. 130

RCOLEC TCH









S.L.86



S.L.82

S.L.92

CENTRAL AVENUE, WEST MOLESEY, SURREY. **TELEPHONE: MOLESEY 4336 (3 LINES)**

D NEW COMPONENTS



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FLEXIBLE SPINDLE COUPLERS, SC.I. Price 2/6.

The lining up of coupled components is made easy with these Polythene insulated flexible couplers. They are completely flexible but entirely free from backlash. The insulator is 14in. dia. and the complete coupler is 4in. in width. The bushes are drilled for 4in. dia. spindle. Obtainable from all reputable stockists or in cases of difficulty direct from works. General catalogue covering technical information on full range of components, 1/- post paid.

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Stop Press: "OSRAM" "912 PLUS" AMPLIFIER

Gold finished front panel with brown control markings Chassis, aluminium, completely punched Pre-Amplifier Chassis, completely punched 6/--

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Chassis, front panel, printed dial plate and drive mechanism, 37/6. Aerial coil O/T1, 2/9. R.F. Coil O/L1, 2/6. Oscillator Coil O/L2,

2/-. 1st, 2nd and 3rd IFT's, IFT 11/10.7, 6/- each. Ratio Discriminator Transformer O/T2 (T5), complete with crystals, 19/6. "MULLARD F.M. TUNER"

MOLLARD F.M. 10NER I.F. Rejectors, Ref. 510/IFF, 2/6. Aerial Coil L1/L2, Ref. 510/RF, 2/6: 4/6. Choke L3, Ref. 510/RFC, 2/-. R.F. Coil L4, Ref. 510/RF, 2/6: Oscillator Coil L5/L6, Ref. 510/OSC., 4/6. 1st LFT L7/L8, Ref. 510/IFT1, 7/6. 2nd IFT L9/10, Ref. 510/IFT2, 7/6. Ratio Detector Transformer L11/12/13, Ref. 510/RDT, 12/6. Aluminium Chassis, completely punched, 12/-.



SALES ORGANISATION A number of progressive appointments will shortly be available. Men of experience and ability who see a future in the field of tape recording are invited to apply for further particulars. (Applications to the Sales Director.)

Products of the Simon Group are the Simon SP/2 Recorder shown left (Simon Sound Service Ltd.), and the Long Duration Tape Monitor LDT. 14 shown right (Simon Equipment Ltd.)



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SIMON SOUND SERVICE LTD. 46/50 GEORGE STREET. LONDON, W.I WELbeck 2371 (5 lines 46

WIRELESS WORLD

NOVEMBER, 1955

an open

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to all manufacturers



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TYPE A 5 VALVES

TYPE A: 5 valves 3-wavebands Superhet with full negative feedback and A.V.C. Built-in Ferrite antenna. Full range tone control. £9/19/6

RADIO RECEIVER CHASSIS

Built to the highest specifications, these chassis offer the finest value to the enthusiast. Supplied with set of selected knobs. Socker panels for aerial, earth, speaker, pick-up and gram motor. 250/50 cycles only. 200

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TYPE B. 7 valves 3-wavebands Superhet with specially designed push-pull output stage. Separate Bass and Treble £15/4/6 control.

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A six-valve pure F.M. chassis with single waveband only, covering all existing and projected B.B.C. FM transmissions. Highest degree of I.F. amplification making it ideally suitable for fringe areas. Output stage specially designed around an EL 41 output valve ensuring a really wide audible frequency range. Permeability-tuned circuit with high stability factor. Special wide-range tone control. Output 4 watts. A.C. 50 cycles only. Provision for external speaker. 13 Gns. Co-axial socket for dipole aerial.

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Packing and carriage for all Chassis 12/6.



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TRUVOX 2-speed, twin track, Tape Deck of the latest type, with push-button £23/2/0 £23/2/0 controls.

Tape Recorder, Amplifier only. Built to the highest standards, magic eye for indicating recording level.

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Cat. No. LS/5-5in. Standard		
Electrostatic Loudspeaker LSH75, for treble response		

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NOVEMBER, 1955

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Write for complete specification of this amazing instrument and literature on the Spectone Magnetic Reproducer and the Spectone 510 Mullard Amplifier.

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NOVEMBER, 1955

WIRELESS WORLD



MARCONI-SIEMENS Radio Telephone Terminal TYPE B (HW 21)

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NOVEMBER, 1955

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Actual Size of the A.V.C. Instrument (Covers removed).

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now available in the sub-miniature 4-stage

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Weighs only 12 ounces complete with battery



why engineers specify EGEN potentiometers –

Egen Potentiometers are based on long experience of requirements of television and electronic equipment manufacturers. In design, dependability, accuracy and freedom from wear they are *outstanding*, but, above all, they are completely NOISELESS.



DUAL POTENTIOMETERS with concentric operating spindles. The new Egen Dual Potentiometers incorporate all these outstanding design features — multiple contact rotors, smooth easy movement, thorough screening between sections, plus a convenient soldering tag for earthing screened connec-

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PRE-SET POTENTIOMETERS. Completely enclosed in high-grade phenolic mouldings. Solder tags heavily silver plated for quick soldering. Fully insulated spindles with integral control knobs. Tapped for 2-hole 6 B.A. fixing on $\frac{3}{2}$ " centres. Type 126, wire-wound. Type 127, carbon.



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component is smooth and noiseless in action and is designed to meet the many and varied requirements of the Electronic Industry. Egen pre-set resistors can be supplied in multi-bank assemblies to suit individual requirements. There are also twin-track models, and types with an electrically divided slider, giving adjustment on two resistors with one operation.

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NOVEMBER, 1955

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STABILISED POWER UNIT





This small, compact, general purpose unit is ideal for laboratory use or production testing where a stable D.C. supply is essential.

input-210, 230, 250 V., 40-100 C.P.S.

Output-Stabilised, 120-400 V., 120-80 mA (in two ranges).

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Stabilisation-Less than 0.15 V. change in output voltage for 10 V. change in mains voltage.

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For use in research and

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(up to 16 channels) to special

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RR SERIES 100

A new high quality Record/Reproducer utilizing separate record, and a transistor playback amplifier with full monitoring facilities during recording.

The transistor head amplifier and equalizing stages of the playback unit permit an extremely low hum level, while the playback quality from this amplifier is of the highest order.

The playback characteristics conform closely to the recommended C.C.I.R. specification, so that recordings made on this instrument, as well as the replay of tape records, require only that the "BASS" and "TREBLE" controls on the main amplifier be maintained in the "FLAT" position, to obtain the correct playback response.

For full information on the Reflectograph Range write to the Manufacturers **RUDMAN DARLINGTON (ELECTRONICS) LTD.** Wednesfield, Staffs. Tel: Wolverhampton 31704 53

NOVEMBER, 1955

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WIRELESS WORLD





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TYPE "B"

14in. dia. × fin. deep. Small ribbed knob in walnut, ivory or black. 4 B.A. grub screw and locknut fixing. Matches type "A.

TYPE 5060

14 in, long, and 3 in, deep. Useful pointer con-trol knob in black with white nose marking. Fixing: 11in. long, 1in. wide, marking. Fixing: grub screw and locknut.

knob in black or white. Brass insert and one 4 B.A. grub screw for fixing.

TYPE 324

14in, dia. × fin.

Medium-sized fluted

deep

TYPE 327

11in. long, fin. wide, 14in. long, §in. wide, 4in. deep. Popular instrument knob in black or white with suitable nose marking. Moulded with brass insert. Fixing: one 4 B.A. grub screw.

TYPE X5/P.1325 11in. dia. × fin. deep. Black fluted instrument knob with brass insert. Fixing: two 2 B.A. grub screws.

TYPE X9/P.1375 lin. dia. \times in. deep. Identical to Type X5/P.1325. Ideal for test apparatus, etc.

TYPE D 2265

14in. long, gin. wide fin. deep. A particularly attractive cream pointer knob with black nose marking, Positive fixing by means of one 4 B.A. grub screw,





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ALL	PRIMARIES	ARE 200/250 v.	Half Shrouded
LICM	42 (Midaat)	Querrue 250 0 250 v	60 mls 63 x 28 32

5 v, at 2 amps,	16/3
HS63. Output 250-0-250 v. 60 m/a., 6.3 v. at 3 amps., 5 v.	
2 amps	16/6
Output.	19/-
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HS2X. 250-0-250 v. 100 m/a., 21/ HS75. 275-0-275 v. 10	00
m/a.	21/-
HS30X. 300-0-300 v. 100 m/a., 21/ HS3X. 350-0-350	V
100 m/a.	21/-
Fully Shrouded	
FSM63 (Midget). Output 250-0-250 v. 60 m/a., 6.3 v. at 3 amp	s.,
5 v. 2 amps.	16/9
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Output.	
ES2 250-0-250 v 80 m/2	21/-
FS30. 300-0-300 v. 80 m/a., 21/ FS3. 350-0-350 v. 80 m/a.	21/- a. 23/-
FS2 X. 250-0-250 v. 100 m/a., 23/ FS75. 275-0-275 v. 100 m/ FS30 X. 300-0-300 v. 100 m/a., 23/ FS3 X. 350-0-350 v. 10	a. 23/-
m/a.	23/-
All the above have 6.3 4-0 v. at 4 amps., 5-4-0 at 2 amps.	
FS43. Output 425-0-425 v. 200 m/a., 6.3 v. 4 amps., C.T. 6.3	v. 47/6
4 amps., C.T. 5 v. 3 amps. Fully shrouded FS50. Output 450-0-450 v. 250 m/a., 6.3 v. 2 ampsi, C.T. 6.3	
4 amos., C.T. 5 v. 3 amos. Fully shrouded	67/6
F35X. Output 350-0-350 v. 250 m/a., 6.3 v. 6 amps., 4 v. 8 amp	S.,
4 v. 3 amps., 0-2-6.3 v. 2 amps. Fully shrouded	65/-
FSI60X. Output 350-0-350 v. 160 m/a., 6.3 v. 6 amps., 6.3 3 amps. 5 v. 3 amps. Fully shrouded.	

 HS150. Output 350-0-350 v. 150 m/a., 6.3 v. 3 amps., C.T. 5 v.

 3 amps. Half shrouded.

 F36. Output 250-0-250 v. 100 m/a., 6.3 v. 6 amps., C.T. 5 v.

 3 amps. Fully shrouded.

 FS120. Output 350-0-350 v. 120 m/a., 6.3 v. 2 amps., C.T. 6.3 v.

 2 amps., C.T. 5 v. 3 amps. Fully shrouded.

 FS150X. Output 350-0-350 v. at 150 m/a., 6.3 v. at 2 amps., C.T. 6.3 v. at 2 amps., C.T. 6.3 v. at 2 amps., T.T. 5 v. at 3 amps. Fully shrouded.

 The above have inputs of 200/250 v.

 29/6 29/9 31/6

OUTPUT TRANSFORMERS

MIDGET OP. 5.000Ω to 3Ω	3/9
8.000Ω to 3Ω	3/9
8,00012 to 312	
OPI0. 10/15 watts output. 20 ratios on Full and Half Primary	17/9
	25/9
Williamson's O.P. Transformer to Author's specification f.	4/13/6
Chokes for Williamson's Amplifier, 30 H. at 20 m/a	16/6
10 H. at 150 m/a	32/-

FILAMENT TRANSFORMERS

A11	2001	230 4.	mpuc	
0/11	E2 M	6 2	01	5

F3. 6.3 v. @ 3 amp. 8/11. F3X. 6.3 v. @ 1.5 amp	5/9
F4. 4 v. @ 2 amp. 7/6. F6. 6.3 v. @ 2 amp	7/6
F6X. 6.3 v. @ 0.3 amp. 5/ F12X. 12 v. @ 1 amp	7/9
FU6. 0-2-4-5-6.3 v. @ 2 amp. 10/ F12. 12.6 v. tapped 6.3 v.	16/6
@ 3 amp F24. 24 v. tapped 12 v. @ 3 amp	23/6
F29. 0-2-4-5-6.3 v. @ 4 amp	17/9
FU12. 0-4-6.3 v. @ 3 amp	17/6
FU24. 0-12-24 v. @ 1 amp	17/6
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Transformers suitable for Low Voltage Lighting. Fully shrouded with terminal blocks, 230 v. Input. 12 v. @ 20 amp. £6. 12 v.	
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F5. 6.3 y. @ 10 amps. or 5 v. @ 10 amps., or 12.6 v. @ 5 amps.,	1101
or 10 y. @ 5 amps	34/
F6/4. Four windings at 6.3 v. tapped 5 v. @ 5 amps. each, giving	
by suitable series and parallel connections up to 6.3 v. @ 20	
amps.	51/6
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A low-cost general purpose amplifier of outstanding performance and distinguished appearance. Output is 5 watts, with matching for 15 ohm and 3 ohm speakers. Input is switched for pickup (90 mV) or microphone (30 mV). In gramophone condition, distortion is 0.8% for 3 watts output at 1 kc, and frequency range is 40/30,000 cycles \pm 3 db, 30 db of negative feedback being RETAIL employed. B.V.A.valves used throughout.





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POWER AMPLIFIER

HF25

REMOTE CONTROL UNIT

HF25A

price complete:

40 GNS

PRNHI

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A

PROVOST AMPLIFIER HF25

26db negative feedback and an output from 2 c.p.s. to 160,000 c.p.s. (over 16 octaves) Practically distortion-free response combined with the infinite damping factor gives a standard of reproduction formerly unobtainable The output transformer is specially designed to meet the exacting specification of the amplifier The amplifier can be controlled from a distance of up to 20 ft. (6m.) Cathode follower output from the remote control unit minimises cable losses.

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Smooth, highly flexible controls and facilities for record player, tape recorder, microphone and radio tuner inputs \bullet Five types of plug-in compensators are available, which match all known types of pick-up \bullet Four switched inputs and a choice of four record replay characteristics for U.S. COL. L.P., R.I.A.A. or EUR. L.P., U.S. 78, or EUR. 78 \bullet Continuously variable lift and cut controls for bass and treble with clearly marked level positions \bullet Treble filter control gives three sharp cut-off frequencies and an unrestricted response position.

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A^N Ediswan Mazda aluminized picture tube gives a picture 60% brighter and more contrasty than is possible with an ordinary tube.

In addition, Ediswan aluminizing protects the screen from ion burn and, with the new Ediswan ion trap tetrode gun to protect the cathode, tube life is increased.

Ediswan production methods, which include the special in-line vacuumizing system, ensure a higher, more uniform standard of lasting efficiency. For complete satisfaction demonstrate and recommend Ediswan Mazda aluminized picture tubes.



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Ediswan aluminized tubes have a mirror backing to the screen. All the light is thus thrown forwards giving brighter, clearer pictures and extra life.

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RV9

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NOVEMBER, 1955

"His Master's Voice" announce the release of the first list of

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revealing to the music lover the amazing achievement of recorded sound in depth and breadth recently demonstrated in the "HIS MASTER'S VOICE" HALL OF SOUND AT THE RADIO SHOW, EARLS COURT.



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STEREOSONIC Tape Recordings

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NOVEMBER, 1955

together on

Aerial

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Simple but revolutionary design principles are incorporated in the sensational new HI.LO range of combined Band 1 and Band 3 aerials—providing an inexpensive yet thoroughly efficient answer to present-day aerial installation problems.

See below the advantages offered by the HI.LO range and the models available.

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NOVEMBER, 1955



An extremely sensitive 5-valve tuner with internal power supply for AC Mains 200/250 volts. 2 position switch for Gram and F.M. (tuner switched off in Gram position). Permeability tuned and temperature compensated against drift. Geared drive with illuminated calibrated dial Valve line-up: ECC85 twin triode as grounded grid RF stage and mixer. 6BJ6 I.F. amplifier 6BJ6 limiter 6AL5 ratio detector 6V4 rectifier. Your existing radio or amplifier con-verted to FM in a few minutes. Price £13 15 0 Complete

(Including Tax) Carriage and Packing 7/6 extra.



Three Wavebands:-

SW 13.6 m. to 50 m.; M.W. 195 m. to 550 m.; L.W. 800 m. to 2,000 m.

latest type valves: 12AH8, 6BW6, Push 6AT6, 12Ax7, 2-6AQ5, 5Y3. Pull Output. AC Mains 200/250. Output transformer fitted. Large Large illuminated dial. Sockets for speaker and pick-up. 4 position wavechange switch. High Q RF and IF colls. Special detector circuit for high gain and superior A.V.C. Complete with knobs.

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The G.E.C. metal cone loudspeaker gives lifelike reproduction of any type of sound over a range of 9 octaves. This includes the whole musical fundamental range with overtones. This gives the true tonal quality and character that all music lovers demand.

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This is a professional instrument but its remarkably low price makes it particularly valuable to the home constructor. It must be used under the correct conditions to obtain optimum results. Cabinets have been specially designed for use with this speaker. Home constructors are invited to write for details.



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TYPE L.O. 352

"L.O.352" IS THE TYPE NUMBER OF AN ENTIRELY NEW ALLEN LINE OUTPUT AUTO-TRANSFORMER NOW AVAILABLE.

Note the following "Star" features:

- ★ E.H.T.: 14 to 18 KV.
- ★ E.H.T. Regulation: Better than 5 MΩ.
- * Audible Whistle: Negligible.
- Application: Self-running, Square-wave or Sawtooth driven.
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For full details please request High Stability Carbon.

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NOVEMBER, 1955



-TUNERS-

The combined V.H.F. Frequency and World Wide Amplitude Modulation Tuner uses a tuned R.F., self oscillating additive mixer, 2 x I.F. amplifiers, and Radio Detector on the V.H.F. band II with completely stable tuning. The AM section uses a tuned R.F. stage, F.C., High gain I.F. amplifier with variable selectivity, and delayed Amplified A.V.C. on all wave bands. A Cathode Ray tuning indicator is fitted and operates on all channels and bands. Available in two types.

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For three years the FM81 has received the acclaim of the trade and private user alike. Completely stable tuning, distortionless output, high AM reject ratio, and tuning indicator are some of the many features that make it the most sought after FM tuning unit to-day! New valves and circuits now make possible the FM81 Mark II. Similar in size and appearance to its famous predecessor but with the following improved features.

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FM8

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The EDWARDS ¹/₄lb. capacity model, illustrated, enables the research engineer to carry out sintering, melting and casting up to a temperature of 1,800°C.

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INDEPENDENT SIDEBAND **RECEIVER Type GFR 552**

... developed to British Post Office specification, and used on their international circuits

The Independent Sideband Receiver type GFR 552 is designed for operation on long-distance, point-topoint, short-wave radio links forming part of the international trunk network. On independent sideband working, the GFR 552 provides facilities for the reception of two single sideband signals, each 6 kc/s wide, one above and one below the frequency of a reduced-level pilot carrier. Each sideband will accommodate either two 3 kc/s wide telephony channels, or several voice frequency telegraph channels. The GFR 552 may also be used for reception of single sideband or double sideband transmission. In the case of the second application this receiver offers two advantages: firstly, the absence of non-linear distortion which occurs in normal d.s.b. receivers when signals are subjected to selective fading conditions; and, secondly, the ability to select upper or lower sideband for demodulation, dependent upon which is freer from adjacent channel interference. The circuit and chassis layout of the GFR 552 closely follow that of the Mullard Receiver GFR 551, which was manufactured for the British Post Office to their design. Special features of the GFR 552 include a high order of oscillator stability and freedom from cross-modulation through which cross-talk between channels or intermodulation between wanted and unwanted signals might occur. A brief technical summary is given below. More detailed information supplied on request.

FREQUENCY RANGE—4-30 Mc/s. NOISE FACTOR—better than 7 dB over the band. SIGNAL TO NOISE RATIO—25 dB for 4 microvolts peak sideband ir put over the band.

SELECTIVITY-The response is flat within 21 dB for sideband frequencies between 100 c/s and 6000 c/s. At 10 kc/s from the requestions of the response is -60 dB relative to the pass band. A.F.C.—The a.f.c. system operates effectively with a pilot carrier level of -26 dB relative to 1 microvolt (which corresponds) to a peak sideband level of I microvolt and a signal to noise ratioof is dB)

NON-LINEAR DISTORTION - Third order intermodulation products which might result in cross talk between sidebands do not exceed 50 dB relative to the sideband levels.

OUTPUT-Variable up to+14 dB relative to ImW into 600 ohms.



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NOVEMBER, 1955

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6A	6 watts	6 only	3/32" (2.4 mm)	0.25 oz.	6"	25/-	1/8
6	6 watts	6 only	1/16" (1.6 mm)	0.25 oz.	6″	25/-	fixed bit

Model II-Special High Temperature Model.

MODEL

A letter from TUNIS!

MODEL

Jean-Michel Tosello, 45 Avenue Jules Ferry, Tunis.

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MODEL

Free Translation.

Dear Sirs, I have been waiting to finish the construction of my Brick Reflex to write and let you know how happy and proud I am of my acquisition. Thanks to the advice which you have been good enough to give me, I have secured an acoustic chain which is in every way remarkable. The results in my apartment are, in fact, superb, instead of the almost deplorable acoustic system which I had previously installed. I congratulate you therefore once again on your marvellous Loudspeakers.





A NEW TECHNIQUE IN HIGH SPEED WAVEFORM MONITORING

BANDWIDTH :

10 kc/s to 300 mc/s

- INPUT IMPEDANCE OF EACH PROBE : Approx. 1 pf (input element of variable capacity divider)
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Provision is made for accurate measurement of time and voltage scales of a waveform

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HIGH SPEED RECURRENT WAVEFORM MONITOR TYPE 500

The wide bandwidth and high sensitivity of the instrument as well as the very high input impedance result from the use of a sampling technique. During each recurrence a measurement is made of the instantaneous amplitude of one point in the waveform. This measurement is amplified and applied to the cathode ray tube as one co-ordinate of a graph of the waveform. During subsequent recurrences, instantaneous measurements are made of different points, resulting, after about 100 recurrences, in a complete graph.

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Lowther FM Mk.II Tuner £22. Plus £7.6s.4d. P.T.

Existing AM/FM tuner units can be modified for A.F.C. Apply for details.

Advanced circuit technique includes • tuned R.F. Pentode stage • separate oscillator valve • Pentode mixer • high gain I.F. stage • limiter stage • FOSTER-SELEY discriminator • Automatic frequency control valve • cathode follower output • 50 c.p.s. injection "check tune."

N.B.—Only Lowther offer all these features.

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No finer sound than that which is broadcast can be attained. Lowther gives you just this.

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do you know...?



that a vibration amplitude of .001" at 500 c/s represents an acceleration of about 25 'g'

The vibratory force required to move a mass load at high frequencies is approximately equal to the inertia force, Mass x Acceleration. Fig. 1 shows the



amplitude response required for vibration testing of aircraft equipment and Fig. 2 shows the peak acceleration curve for the same conditions. For a total mass of, say, 10 lb., it can be seen that the force requirements increase enormously as the frequency is increased. A vibration generator with a force output of at least 250 lb. would be necessary to perform the tests at 500 c/s and, of course, the same unit can be used for the lower force output.

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NOVEMBER, 1955

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The latest Miniature Earphone is characterised by its high efficiency and slim, elegant appearance. The curve shown above was taken on a Post Office type 11 c.c. artificial ear and the dotted lines represent the 95% confidence limits calculated from a batch of Earphones of this type.

Controlled methods of manufacture, in which the magnet in each unit is individually adjusted and aged; plus a detailed inspection procedure which includes an automatic curve trace on each Earphone, ensures a uniform product of high quality and reliability.

 λ wide choice of impedance values can be produced to meet the designer's needs.

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TAPE RECORDER COMPONENTS AND ACCESSORIES

TAPE DECKS MARK IIIU SERIES

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Wireless World

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RADIO, ELECTRONICS, TELEVISION

Managing Editor: HUGH S. POCOCK, M.I.E.E. Editor: H. F. SMITH

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NOVEMBER, 1955



ALVES, TUBES & CIRC

35. HIGH QUALITY SOUND REPRODUCTION WIDE RANGE OF CIRCUITS

The high quality reproduction obtainable with modern audio valves at low cost has already been demonstrated by the circuit for a 5-valve 10-watt amplifier introduced last year. The '5-10' is now a firm favourite because of the easc with which it can be constructed and because it can be driven directly by many of the more popular and inexpensive types of crystal pickup.

Many constructors will, however, be interested in building a preamplifier for the '5-10' in order to use a pickup giving 40mV output, perhaps a magnetic type or a crystal pickup loaded for output proportional to stylus velocity. Suitable single-valve units of straightforward construction have been designed by Mullard engineers. There are two versions of the basic design. The first is referred to as preamplifier 'A', and is the one required by those who have already made a '5-10' in which the treble, bass, and volume controls are included in the main amplifier. The second version is known as preamplifier 'B', and is more suitable for those who are only just going to build the amplifier. Here the tone and volume controls are included in the preamplifier itself, giving more convenient access to the controls-particularly important if the equipment is to be assembled into a cabinet. The tone and volume controls in preamplifier 'B' are similar to those which can be incorporated in the main amplifier (as in the original design) but with the appropriate changes to component values.

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Both 'A' and 'B' are suitable for microphone input, or tape input with 30dB bass pre-emphasis at 120c/s.

Point-to-point wiring diagrams and chassis drawings are given in the booklet 'High Quality Sound Reproduction' for preamplifiers 'A' and 'B' and for the two corresponding versions of the '5-10' itself. The booklet also gives details of minor improvements and modifications made to the original amplifier circuit. Chiefly they are intended to take advantage of the high quality of the output transformers which have become available commercially and to give increased stability at the high and low frequency ends of the response. The rated output of the amplifier remains at 10 watts, but the overload point (onset of clipping with



NEW PUBLICATIONS

'High Quality Sound Reproduction' is available from radio retailers price 3s. 6d. It contains 48 pages, of which 44 are devoted to the designs described elsewhere on this page. There are 24 pages of information not previously published, including 14 pages of constructional drawings. The pages are 10 ins. by 8 ins.

The 'Mullard Maintenance Manual' is now on sale at radio retailers price 10s. 6d. It provides the information of most use to the service engineer for repairing and maintaining radio and television sets and audio amplifiers, and includes data on valves and tubes for maintenance purposes as well as on the latest types. Constructors and designers will also find this manual extremely useful as a quick reference guide to data. For the complete information required in designing new equipment the Mullard Technical Handbook (available on a subscription basis) should be consulted.

To help home constructors starting to experiment with transistor circuits a new leaflet is available free of charge entitled 'Junction Transistors for the Home Constructor'. It contains eight circuits using Mullard OC70 and OC71 junction transistors, which are now readily available. Background information is included for those who would like to go more deeply into the subject. normal or low loading is employed in the output stage. The EZ81 has the same pinning as the EZ80 originally specified, but the *total* limiting resistance at each anode of the EZ81 must be at least 310Ω in this circuit.

Other Mullard circuits designed for high quality sound reproduction have been published recently in Wireless World. A 20-watt amplifier, sufficient for a small to medium-sized hall, was described in the May and June issues. The use of distributed load operating conditions in the EL34 push-pull output stage gives a total harmonic distortion of less than 0.05% at the 20 watts nominal output. A 3-valve preamplifier suitable for this and similar amplifiers was described in the July Wireless World, and the August issue contained an article on an f.m. tuner unit which can provide the input for the 10-watt or 20-watt amplifier. Point-to-point wiring diagrams and chassis drawings for these circuits, in addition to reprints of the Wireless World articles, are included in 'High Quality.Sound **Reproduction**'.

A 3-valve 3-watt amplifier has been designed for those who re-

quirereasonably high quality at low cost and with the simplest possible construction. The total harmonic distortion is 1.5% at 3 watts; the sensitivity is 100mV, and the amplifier is suitable for all types of crystal pickup. This amplifier can be driven by the f.m. tuner unit via a 5:1 attenuator. An EZ81 must be included to provide power for the f.m. tuner, and the mains transformer must then be rated at 90mA or 100mA. Preamplifier 'A', primarily designed for the 10-watt amplifier, is suitable here also; a 5:1 attenuator is required to preserve the original basic sensitivity. Without the attenuator the sensitivities are increased by a factor of 5. Details of the 3-valve 3-watt amplifier are available only in the form of a leaflet (free, from the address below).



The free leaflets and further information can be obtained from

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* Very compliant stylus-superb transient response, low record and sapphire wear.

Sapphire styli replaced very simply without any tools whatsoever.

Positive turnover mechanism with a neutral position.

Precision sapphires made by the finest precious stone lapidiaries. Individually inspected under 500x magnification.

Each unit very carefully checked for sensitivity and response and finally subjected to a critical listening test before being despatched from the factory.



ACOS devices are protected by patents, patent applications and registered designs in Great Britain and abroad. COSMOCORD LIMITED · ENFIELD · MIDDX · TEL: ENFIELD 4022



By the time this reaches the reader we hope that the G9AED test card will have been recognised by many hundreds of new viewers, and we do want everybody to know that every report received will be appreciated, and will be acknowledged by Q.S.L. card. Early reports are the most valuable, but please do not telephone, and please send reports to Enfield and not to the transmitter.

The transmitting equipment has been installed in the large trailer illustrated above, the power unit is a separate self-controlled 5 kW generator.

Regular transmissions should commence at 10 a.m. on Monday October 10th, and will be as follows:—Monday to Friday 10 a.m. -1 p.m., 3 p.m.-6 p.m. and 7.30 p.m. -8.30 p.m., Saturday 10.0 a.m. to 1 p.m. There will be no transmissions on Sundays or Bank Holidays.

From the results of our experience on London, we can tell you that it is a reasonable assumption that a locked picture of G9AED, in which all characters can be read, will secure a first class result for I.T.A. when it commences, and without any further major adjustment either to the aerial or receiver. A barely discernible and completely unlocked picture from G9AED will result in a picture of moderate entertainment value from I.T.A., but there will in all probability be some noise present on the picture. Please remember that a basis of comparison will exist, and viewers will expect to receive the I.T.A. programme as well as they have been accustomed to receive the B.B.C., and without constant readjustment of set controls.

Where indoor aerials are being tried, great care must be taken in siting as the problem of standing waves is very real. Correctly designed aerials for loft use are quite satisfactory when used in situations of strong signal.

INSULATED COAXIAL PLUG L781/P2

This plug is in every way interchangeable with our well known all-metal plug L734/P/AL. It is not generally known that the drive behind the design of this plug came from certain engineers of the British Electricity Authority who were unhappy regarding the safety aspect in the event of the breakdown of the isolating capacitor in the receiver. Provided that the receiver has been connected to the mains with a threepin plug correctly wired through all stages, there is no possibility of danger of shock. If a two-pin plug and socket has been used, there is of course a 50% chance of connecting it the wrong way round. Even with



a three-pin plug, this might be wrongly wired at any one of at least three points:

(1) The socket (unlikely if competently installed). (2) The plug if it has been replaced by an incompetent amateur, or (3) If there has been an error within the receiver.

The insulator is not a brittle moulding, it is nylon, and even if stood on will come to no harm. It is more robust than those manufactured in metal.

> Advertisement of BELLING & LEE LTD. Great Cambridge Rd., Enfield. Middx. Written 24th Sept., 1955



HEAVY CURRENT LOW VOLTAGE FUSE LINK

for use on D.C. supplies in Aircraft, Battery Driven Vehicles, &c.



THE fuse is intended as a direct replacement for the American "Current Limiter" used on

28V d.c. aircraft power supplies, and for this purpose has received the approval of the Air Registration Board under their reference No. E.3217.

It is also suitable for the protection of battery operated vehicles, heavy current rectifier output circuits, low voltage furnaces and other similar d.c. applications.

SPECIFICATION: The fuselink is capable of breaking currents of up to 3,000 amperes at 30V d.c. (2,500 amperes at 30V d.c. for the 35A and 50A ratings) without any external disturbance. The arcing time does not exceed 0.003 seconds at 3,000 amperes.

RATINGS: 35A, 50A, 80A, 100A, 130A, 150A, 200A, 225A, and 275A. The range will be extended in four further steps to 500A maximum.







Marconi Surveying Service

Before planning any communication system, and particularly a microwave or V.H.F. multichannel system, a survey of the propagation conditions over the proposed path or area is essential. Similar, but less exhaustive surveys, are also necessary before planning V.H.F. mobile systems. Such surveys are undertaken by Marconi's, one of the very few radio manufacturers who do so. The teams engaged in the work may be called upon to operate in desert, swamp and jungle, over which line and cable routes would be impractical, on windswept moorlands or in densely populated city and suburban areas. Surveys are being, or have already been carried out all over the world, including : Uganda, Kenya, Tanganyika, Nigeria, Gold Coast, Tangier, Azores Norway, Turkey, Greece, Malaya, Ceylon, West Indies, Sweden, and also, of course, in Britain.

LEFT. Balloon operations on the Ipoh-Telok Anson route in Malaya.

RIGHT. The mast is up and the motor generator is running during the survey of the Nigerian multichannel system.

BELOW. The V.H.F. mobile survey team erect their masz.

LC 10

Over 80 countries now have Marconi-equipped telegraph and communications services. Many of these are still giving trouble-free service after more than twenty years in operation.

Lifeline of communication

MARCONI

COMPLETE COMMUNICATION SYSTEMS Surveyed, planned, Installed, maintained

MARCONI'S WIRELESS TELEGRAPH CO., LTD., CHELMSFORD, ESSEX

EMITAPE 88 in ever-increasing demand

Emitape is used by the great recording companies - "His Master's Voice ", Columbia and Parlophone - and by the world's leading broadcasting organisations, because of its fidelity, sensitivity and length of life. This easy-to-use tape is also being increasingly employed in laboratories, factories and domestically - in fact, wherever true-to-life recording is required. It is made in a variety of spool lengths wound on plastic or aluminium spools to meet the requirements of professional and domestic recorders with differing hubs.

Special Features

- . HIGH SENSITIVITY . HIGH TENSILE STRENGTH
- P.V.C. BASE ANTI-STATIC
- FREEDOM FROM CURL
- . EDITING LEADER AND TRAILER STRIPS

TAPE ACCESSORIES

- 7" EMPTY PLASTIC SPOOL IN CARTON AP.87 5/6
- 5" EMPTY PLASTIC SPOOL IN CARTON AP.85 4/6
- 3" EMPTY PLASTIC SPOOL IN CARTON AP.93 3/-

Non-Magnetic Scissors AP.39 Price 16/-



AP.46 Price 8/-

Magnetic Tape Jointing Block

175 ft . . . 3 in.

PLASTIC SPOOL

(88) . . . 716



608 ft . . . 5 IR.

PLASTIC SPOOL

(88 6) ... 21 -







Gummed Jointing Tape AP.37 Price 6/6

1200 ft . . . 7 In. PLASTIC SPOOL

(88 12) ... 35 -

ENUITADE



Used by the leading Broadcasting Organisations and for recording the world's greatest artists on

'HIS MASTER'S VOICE,' COLUMBIA AND PARLOPHONE RECORDS



Full particulars of Emitape and editing accessories together with literature are obtainable at your local dealer or from:---Telephone: SOUTHALL 2468 E. M. J. SALES & SERVICE LTD. RECORDING EQUIPMENT DIVISION, HAYES, MIDDLESEX. Expert Enguiries for products mentioned in this advertisement should be addressed to :

Brandard

The modern trend is towards super-highfrequency radio links for broad-band communications, a field in which *Standard* has played a leading role since 1931. Including recent orders *Standard* SHF links will effectively contribute a potential capacity of no less than 5 000 000 telephone channel miles to world communications. The advantages of *Standard* SHF Radio links, particularly over terrain where cable laying presents difficulties, are that each installation can provide up to 7 both-way radio channels. Each channel may be equipped to carry television signals of 405-, 525- or 625-line definition standards or up to approximately 600 telephone channels.

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Standard Telephones and Cables Limited Registered Office : Connaught House, Aldwych, London, W.C.2

RANSMISSION DIVISION, NORTH WOOLWICH, LONDON, E.I.

VORTEXION



The amplifier, speaker and case, with detachable lid, measures 81 in, x 221 in, x 153 in, and weighs 30 lb.

PRICE, complete with WEARITE TAPE DECK .. £84 0 0 TAPE RECORDER

* The total hum and noise at 74 inches per second 50-12,000 c.p.s. unweighted is better than 50 dbs.

The meter fitted for reading signal level will also read bias voltage to enable a level response to be obtained under all circumstances. A control is provided for bias adjustment to compensate low mains or ageing valves.

* A lower bias lifts the treble response and increases distortion. A high bias attenuates the treble and reduces distortion. The normal setting is inscribed for each instrument.

+ The distortion of the recording amplifier under recording conditions is too low to be accurately measured and is negligible.

* A heavy mu-metal shielded microphone transformer is built in for 15-30 ohms balanced and screened line, and requires only 7 micro-volts approximately to fully .oad. This is equivalent to 20ft. from a ribbon microphone and the cable may be extended 440 yds. without appreciable loss. The .5 megohm input is fully loaded by 18 millivolts and is suitable for crystal P.U.s, microphone or radio inputs.

A power plug is provided for a radio feeder unit, etc. Variable bass and treble controls are fitted for control of the play back signal.

The power output is 3.5 watts heavily damped by negative feedback and an oval internal speaker is built in for monitoring purposes.

★ The play back amplifier may be used as a microphone or gramophone amplifier separately or whilst recording is being made. The unit may be left running on record or play back, even with 1,750ft. reels, with the lid closed.

POWER SUPPLY UNIT to work from 12 volt Battery with an output of 230 v., 120 watts, 50 cycles within 1%. Suppressed for use with Tape Recorder. **PRICE £18 0 0.**

We supply and recommend the Jason F.M. Feeder Unit. PRICE £15 17 0, including Purchase Tax.

FOUR CHANNEL ELECTRONIC MIXER

is almost essential for the professional or semiprofessional where a number of different items have to be mixed on one tape recording.

It is recommended by a number of tape recorder manufacturers for this purpose.

Any normal input impedance can be supplied to order, balanced or unbalanced, the standard being 15-30 ohms balanced.

The normal output is 0.5 volt on 20,000 ohms or less. but 600 ohms is available as an alternative.

The steel stove enamelled case is polished and fitted with an engraved white panel suitable for making temporary pencil notes. An internal screened power pack and selenium

rectifier feed the five low noise non-microphonic valves.

Used in many hundreds of large public address installations and recording studios throughout the world.



PRICE £36 15 0.

Manufactured by

VORTEXION LIMITED, 257-263, The Broadway, Wimbledon, London, S.W.19

Telephones: LIBerty 2814 and 6242-3

Telegrams: "Vortexion, Wimble, London."

NOVEMBER, 1955 WIRELESS WORLD 98 EMI RECORD DIVISION EXPANSION PLANS

PROGRESSIVE OPPORTUNITIES FOR A WIDE RANGE OF TECHNICIANS & ENGINEERS

The £3,000,000 expansion plans for the Record Division of the E.M.I. Group entails the creation of complete teams of technicians and engineers with varying degrees of experience and capacity. Exceptional opportunities exist for personnel who within a short time show the initiative and capabilities fitting them for responsibilities connected with pioneering new projects.

The vast resources of the E.M.I. Group provide the ideal environment for progressive technicians.

Recently built laboratories have been fitted with modern equipment and plant, and the Company's Recording Studios are the largest in the United Kingdom.

THE RECORD DIVISION of

ELECTRIC & MUSICAL INDUSTRIES LTD ("His Master's Voice", Columbia, Parlophone, Marconiphone)

This will provide excellent opportunities for suitably qualified men in the following capacities.

Immediate requirements :--

(a) RESEARCH & DEVELOPMENT

ENGINEERS AND PHYSICISTS, with a degree or equivalent qualification in chemical, mechanical, and electrical engineering, or physics, for advanced development and design of all types of equipment in the fields of sound recording and reproduction, the physical chemistry of materials used for magnetic and disc recording, and the manufacture of commercial records. Personnel with several years' experience are particularly required, but posts are available for recently graduated men with no National Service commitments.

JUNIOR ENGINEERS and Technicians for work in the same field.

(b) RECORDING STUDIOS RECORDING OPERATORS, with experience of tape recording and studio control work. Also disc recording operators, with experience of modern lacquer cutting.

EDITORS with experience of tape recording and a good musical knowledge.

MAINTENANCE ENGINEERS with experience of modern microphones, amplifiers and studio practice.

(c) POST-GRADUATE TRAINING SCHEME Graduates in physics or electrical, mechanical or chemical engineering with no National Service commitments can be accepted for one-year training

scheme covering all branches of the work of the Record Division, including Recording Studios, Record Processing and manufacture, Development Laboratories and Design Departments, manufacture of recording materials and Tape Records. Men completing this course will be offered technical posts in the Division suitable to their qualifications and aptitudes.

(d) TECHNICAL INFORMATION OFFICER

A man is required with a scientific training, and with good editorial ability to prepare technical reports. Must be able to write lucidly and to be familiar with current sources of technical literature in communication engineering, physics and chemistry. Previous experience in sound recording would be an advantage.

In all the above positions, a knowledge of one or more modern languages would be an advantage.

Successful applicants will be based either at St. John's Wood, London, or Hayes, Middlesex, and may be expected to travel abroad from time to time. All staff over 21 are eligible for membership of the Company's Pension Scheme. Sports facilities and social clubs covering all interests are available. Write for an appointment for interview, giving full particulars of experience and qualifications to :-

MR. H. A. M. CLARK,

TECHNICAL MANAGER, RECORD DIVISION, Dept. E, ELECTRIC & MUSICAL INDUSTRIES LIMITED, BLYTH ROAD, HAYES, MIDDLESEX.



Electric & Musical Industries Ltd. is a famous British organisation concerned with the application, development and manufacture of electronic equipment for industrial and domestic purposes.

Many of its products carry trade marks which are household names including "His Master's Voice", Columbia, Parlophone, and Marconiphone.





SCOTCH BOY **Electrical** Tapes NOW with Thermosetting Adhesive

In the field of component assembly, 'Scotch Boy' electrical tapes have long been unrivalled for strength, ease of application, and excellent dielectric properties.

Now a new range of 'Scotch Boy' electrical tapes, with thermosetting adhesive, has been introduced. These remarkable new paper, glass cloth, and acetate cloth tapes have the same ability to stick at a touch, but the adhesive cures firm when components are subjected to the normal drying cycles.

The cured adhesive has greatly increased solvent resistance, and soft spots are eliminated. The new tapes are, therefore, ideal for use with solventless varnishes and casting resins.





HERE ARE TYPICAL APPLICATIONS :

TOP LEFT	No. 38 Paper Tape binds and insulates motor field coils
TOP RIGHT	No. 38 Paper Tape is used to anchor, start, and finish wires
CENTRE RIGHT	No. 27 Glass Cloth Tape holds the leads and insulates the lead wire splice of a lighting transformer
BOTTOM RIGHT	No. 28 Acetate Cloth Tape anchors leads in second- ary windings, and No. 38 Paper Tape holds fibre lead pads of a transformer

FOR SPEED AND ECONOMY IN COIL ASSEMBLY SCOTCH BO (Regd. Trade Mark) Electrical Tapes

MINNESOTA MINING AND MANUFACTURING COMPANY LIMITED . LONDON . BIRMINGHAM . MANCHESTER . GLASGOW

ANOTHER

PRODUCT

Why Ediswan Clix P.T.F.E. Valveholders are widely used in B.B.C. Television equipment

Large quantities of Ediswan Clix P.T.F.E. Valveholders are used in B.B.C. Television equipment. Only the combination of the finest insulation— P.T.F.E., the most efficient contact material— Berylium copper—and Ediswan Clix design and manufacture can match the requirements of efficiency and reliability in this and all other stringent valveholder applications.

Ediswan Clix P.T.F.E. Valveholders are fully type approved for Services Grade 1, Class 1 conditions. Full details of these valveholders and other components in the Ediswan range are given in catalogue CR. 1681. Manufacturers and Development Groups may have a copy on request.



RADIO, TELEVISION & ELECTRONIC COMPONENTS

THE EDISON SWAN ELECTRIC COMPANY LIMITED, Member of the A.E.I. Group of Companies 155 Charing Cross Road, London, W.C.2 and Branches. Telephone: Gerrard 8660. Telegrams: Ediswan, Westcent, Londo CR3

CURTAIN UP on the "WARWICK"

The "Warwick" may be the first Grundig fm radiogram you've seen; it's a fine instrument.

Incorporating the famous "Kenilworth" chassis it covers all broadcast wave-bands, including V.H.F. There are two built-in aerials, push-button wave change, magic eye tuning and the Duplex. Control station selection that's one of the most ingenious features of Grundig design.

The gram side, as you would expect, has three-speed, mixed record auto change.

Reproduction? It's magnificent. Four speakers. Four compensating speakers to give the sound depth and reality.

Just to round things off there's the cabinet; gilt inlaid -very handsome-and ample built-in record accommodation that lights up automatically as the compartment is opened.

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new radiogram GRUNDIG



WESTON panel instruments

Both round and rectangular models of moving iron, moving coil, A.C. rectifier and H.F. thermocouple types are offered. In the range of rectangular instruments, which have been introduced to give the advantage of long, easily-read scales and to harmonize with rectangular panels, certain models are available with illuminated dials. Full particulars of types and ranges available are to be found in leaflets List Nos. W.1 and W.2, copies of which are available on request.

> Larger instruments, both round and rectangular and for switchboard or panel mounting are also available. These have scale lengths of 6" and 61" respectively.

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Rectangular panel instruments are available with scale lengths of 2.5°, 3.2°, and 4.2°. These offer the advantage of an increase in scale length of approximately 50% over their equivalent round models, for which they can be used as direct replacements using the same panel fixing holes



Round models are housed in cases of 2^{*} , $2\frac{1}{9}^{*}$ and $3\frac{1}{9}^{*}$ diameter and have scale lengths of 1.7^{*} , 2.1^{*} and 2.8^{*} respectively.

101

Newcastle-on-Tyne, Newcastle 26867 Nottingham, Nottingham 42403

102

high efficiency

E long life



£12-10s.

Add to the wide response already obtainable from the famous Phase Inverter Speaker by merely adding the new "Extratop" Dynamic Pressure Tweeter Unit complete with Crossover. Impedance 15 ohms. Suitable for use in conjunction with any 15 ohm speaker but even better with the Phase Inverter Speaker. Tweeter £12 10s; Phase Inverter Speaker £16 10s.





miniaturisation

Ediswan now offer a complete range of Germanium Point Contact and Junction Rectifiers. The former find many applications in the high frequency field and the latter are ideally suited for power packs and similar applications where high efficiency and minimum space requirements are of prime importance.

A descriptive brochure is available on request and our Engineers are at your service.

EDISWAN

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155 Charing Cross Rd., London, W.C.2 and Branches

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TRADITION

YOUNG INDUSTRY IN

The oldest high fidelity amplifiers in the world are of LEAK manufacture. In 1945 as the result of war-time research in our laboratory we were able to offer, to an astonished world of audio engineers, amplifiers with a distortion content as low as 0.1%. A survey of engineering literature will confirm that we were the first manufacturers in the world to design and market amplifiers with such a small distortion content, and the magnitude of this advance can be gauged when it is remembered that the then accepted standard for laboratory amplifiers was 2% distortion. Our figure of 0.1% was received with incredulity, but it was subsequently confirmed by the National Physical Laboratory and this criterion is still an accepted world-wide standard.

With this clear lead on low-distortion amplifiers we were able to build up an export market much greater than the domestic one, and the increased volume of manufacture resulted in lower prices, which, in turn, brought real high fidelity amplifiers within the reach of the music-lover at home.

We have devoted 21 years entirely to the development and manufacture of audio products and we are proud of our position as the leaders in this field. We are also proud of the fact that the "Point One" amplifier supplied to our first customers are still giving them results which, even now, cannot be surpassed. Our research and development departments are ever active, our pre-amplifiers have been re-designed for use with the latest input devices, and we have made great progress in the war on prices. From long experience, by the employment of new techniques and by extreme attention to design details during development work on the pre-production models, we enable our labour force to achieve a high output per man-hour. The labour costs thus saved offset the increased costs incurred for high-grade materials, components and finishes, and this together with quantity production (made possible only by a world-wide market) explains how quality products may be sold at reasonable prices.

To our old customers we give our thanks for their support and recommendation-the basis on which our Company has grown. Those who are seeking to obtain the highest quality of gramophone and radio reproduction would be wise to hear and inspect LEAK products which, with their tradition of excellence, represent the best that can, be obtained.



LEAK TL/IO AMPLIFIER **'POINT ONE' PRE-AMPLIFIER**

"POINT ONE" PRE-AMPLIFIER

The handsome gold escutcheon plate contributes to the elegant appearance, and blends with all woods.

+ Pickup The pre-amplifier will operate from any pickup generally available in the world. A continuously variable input attenuator at the rear of the pre-amplifier permits the instantaneous use of crystal, moving-iron and moving-coil pickups.

Radio The radio input sockets at the rear permit the connection of the LEAK V.S. tuner unit. An input attenuator is fitted. H.T. and filament supplies are available from the pre-amplifier.

+ Distortion

ELECTROSTATIC LOUDSPEAKERS

Reprints of "The Gramophone" article (May, 1955), by H. J. Leak, summarising his work and findings on Electrostatic and Dynamic Loudspeakers, are available on request, free of charge.

27 gns. complete

Hum Negligible, due to the use of recently developed valves and special techniques. And of the selector Radio, tape, records; any and all records can be accurately equalised.

★ Treble Continuously variable, +9 db to — 15 db at 10,000 c/s.

* Bass Continuously variable + 12 db to - 13 db

t 40 c/s. ★ Volume Control and Switch The switch controls the power supply to the TL/10 power amplifier.

to the L/10 power amplither, **★ Tape Recording Jacks** An exclusive feature. Readily accessible jacks are provided on the front panel for instantaneous use with Tape Recorders which have built-in (low level) amplifiers.

TL/10 POWER AMPLIFIER

Circuitry

A triple loop feedback circuit based on the famous TL/12. The output transformer is the same size as in the TL/12.

Maximum power output: 10 watts.

Frequency Response: ±1 db 20 c/s to 20,000 c/s. Harmonic Distortion: 0.1%,1,000 c/s, 7.5 watts output. Feedback Magnitude: 26 db, main loop.

Damping Factor: 25.

Hum: --- 80 db referred to 10 watts. Loudspeaker Impedances: 16 ohms, 8 ohms, and 4 ohms,

★ Write for leaflet W ★

Phone: SHEpherds Bush 1173/4/5

H. J. LEAK & CO. LTD., BRUNEL ROAD, WESTWAY FACTORY ESTATE, ACTON, W.3 Telegrams: Sinusoidal, Ealux, London

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NOVEMBER, 1955

MARCONI Vacuum Tube Voltmeter TYPE TF 1041

> FREQUENCY RANGE 20 c/s to 700 Mc/s AND D.C.

An accurate and most stable instrument for the direct measurement of unbalanced a.c. voltages up to 300 volts, balanced or unbalanced d.c. voltages up to 1000 volts, and resistances from 0.2 ohm to 500 M Ω .

Multiplier units are available as optional accessories and extend the measurement ranges to 3 kV for a.c. and 30 kV for d.c.

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MORE MONEY-SAVING LASKY BARGAINS **O**N 105

PAGE

NEXT

NOVEMBER, 1955





All

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Components

either above Amplifiers supplied separately, for printed circuit or con-

Price Lists on request.

construction.

printed circuit or

and I UAF42. Input voltage 100/100 AC/DC Very easily con-supplied with circuit diagram and all details. Size 9X4X4 in. Uses two metal rectifiers, one each RM2 and RM3. Ideal for ships, record players, tape recorders, home record players, tape recorders, home record players, tape the fully assembled and wired, with four values. four valves. LASKY'S PRICE 65/-, Carriage free.

LASKY'S 4-WATT A.C. AMPLIFIER KIT Uses 1 each 6SL7, 6V6, 5Z4. All components, chassis, valves, output trans., mains trans., £4/5/-.

Carriage and packing 2/6. **INSTRUCTION BOOK and shopping** list, 1/- post free.

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6¹/₄in...17/6. 8in...19/6. 10in ...19/-Plessy H.D. 10in...25/- 6×4 Elliptical 18/6. Plessy 12in...37/6, Post extra.

LATEST COLLARO RC.54 3-speed High Fidelity Mixer Changer, Studio crystal turnover p.u., in leather-ette covered carrying case. £13/5/-.

RECORD PLAYING UNITS 3-speed, auto and hand change. All types in stock—Garrard, Collaro, B.S.R., etc.

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COLLARO TAKE DECK MOTORS Anti-clockwise, shaded pole. LASKY'S PRICE 25/-, post extra.

EX-GOVT. ACCUMULATORS. 2 volt, 10 a.h. Size 1²in. square × 5¹/₂in. high. Made by Canadian Exide. LASKY'S PRICE 4/6. Post 1/-.



5-VALVE RADIO CHASSIS

5-VALVE KADIO CHASSIS Brand new and unused. A.C./D.C. 200/250 volts. I.F. 465 kc/s, A.V.C. 4 watts output, 3-station pre-set, frame aerial, fully aligned, chassis 10X54in, max. height 54in. Completely wired and ready for use, with the addition of a speaker and output transformer. Two controls, volume and station switch. Valves used: 10C1, 10F9, or UF41, 10LD11, 10F14, U404 or UY41! LASKK'S PRICE 69/6 less valves. Post 3/6 extra. With valves £5/19/6.

ORDER BY POST IF YOU CANNOT SAVE POUNDS! CALL

FAMOUS AMPLIFIERS BUILT ON T.C.C. PRINTED CIRCUITS

The latest advance in Amplifier design. We can now supply Inclatest advance in Ampiner design, we can now supply from stock two famous Amplifiers, the Osram 912 and Mullard 5/10, built on the new printed circuit technique. All specified components, T.C.C. condensers, Lab. resistors, etc., are used and you have your choice of transformers and chokes by Partridge, Haddon, W/B or Ellison. Demonstrations given any time.

> The MULLARD 510 AMPLI-FIER, built on T.C.C. printed circuit, supplied fully assembled complete with valves, ready for use. Price, depending on make of transformers used, 15 gns.

> > Printed Circuit separately 22/6

New Mullard Amplifier Book, 3/6.

The OSRAM 912 AMPLIFIER, built on T.C.C. with valves, ready for use. Price, depending on make of transformers used. 19 gns. Printed Circuit separately, 50/-.

Book of the Osram 912, price 4/-.

DRILLED CHASSIS AND DIAL ASSEMBLY Size $13\frac{1}{2} \times 7 \times 2\frac{1}{2}$ in., drilled for five latest type miniature valves, mains trans., I.F., etc. Dial 13×14in., for horizontal or vertical mounting. Spin wheel tuning. All pulleys and spindle supplied. LASKY'S PRICE 19/6 Post 3/-

ALUMINIUM CHASSIS $\begin{array}{c} \text{ABG: underliked, 4 sides, reinforced corners. Depth $2\frac{1}{2}$ in, 6×4 4/-$ 12\times8$ 7/-$ 16\times10$ 8/3$ 8×6 5/-$ 14\times9$ 7/6$ 12\times3$ 4/9$ 10\times7$ 6/-$ 16\times9$ 8/-$ 12\times6$ 6/6$ \end{array}$ 10×10 0/1 12× 3 4/9 12× 6 6/6 Post 1/- per chassis extra

DULCI RADIO CHASSIS Full range 3 and 6 wave, £6/19/6 to, 21 gns.

GANG CONDENSERS 0005, less trimmers.

2-gang, standard, 5/6, min., 6/6. 3-gang, standard, 7/6, min., 10/6. 4-gang, standard, 10/6. With Trimmers : 2-gang, standard, 7/11, min., 7/6. Post extra.

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Deck with space for amplifier, feeder unit and speaker. Overall dimensions with lid closed, 19in.×14in.×13in. LASKY'S PRICE 59/6Carriage 5/-.

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EDITOR HI-FI (illustrated above).

Fitted in elegant leathercloth suitcase with gilt fittings. Complete with High Fidelity desk microphone and 1200 ft. spool of Emitape.

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Fitted in padded simulated crocodile case with continental gilt fittings and locks, with elegantly styled super tape deck. Completely automatic simple interlocked control; MIXING and MONI-TORING facilities. Complete with microphone and spool of Emitape.

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The unique lightweight custom built chassis of the Editor Hi-Fi and Editor Super Hi-Fi embraces the latest techniques of recorder construction.

At this time of the year, we have a look round our warehouses and any items we wish to clear are offered at very low prices. Here is a selection — there are very few of each — so order now to secure a bargain !

AMPLIFIERS Lowther A15F Rogers Baby de Luxe with pre-amp	£31. 19. 6. £19. 19. 0.	PICKUPS, Ronette lightweight. High Fidelity with 2 heads 3 POTENTIOMETERS, 1 ohm., 1 watt
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BALLAST RESISTORS C9266, K55H, K55B,		SOLDER GUN BITS, Burgoyne 12 for
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CABINETS Record Player (GU4) Polished	£2. 2. 0.	TAPE RECORDER AMPLIFIER CONTROL PANELS
DIALS, Ivorine MW, LW, 5W,	2/3	TAPE OSCILLATOR COILS
FILTERS, Coloured T.V. 12" or 15"	5/-	TONE CONTROLS, Goodsell F/TC £7. 19.
HEADPHONES, Type K, S.G. Brown	£4. 4. 0.	TRANSMITTING CONDENSER, 100 pfd. variable 3
.F. TRANSFORMERS, 465 Kc/s per pr.	5/-	TORCHES, 2 cell focussing, adjustable, Type C
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21	Garrard 301	£25	3	6
9	Connoisseur 3-speed	£27	2	6
24	Collaro 2010	£18	11	11
4	Lenco 4-speed	£20	15	5

AUTO-CHANGERS

9	Garrard RC110																	£13	19	7
7	Garrard RCIII									•	• •					•		£14	8	0
9	Collaro RC54	• •	•	•	•	•	•	•	•	•	• •		•	•	•	•	•	£13	9	6

AMPLIFIERS

3	Goodsell PFA	£20	0	0
3	Goodsell Williamson "C" Core	£40	0	0
6	Pye PF91 and 91A	£42	0	0
3	Leak TL12	£28	7	0
3	Leak Varislope Mark II	£16	16	0
10	Leak TL10	£17	17	0
7	Leak Point one pre-amp	£10	10	0
7	Goodsell MAS/UL	£14	17	6
5	Pam 1001 pre-amp	£42	0	0
7	Goodsell ULF TC	£12	12	0
8	Quad amp. and control unit	£42	0	0
3	Lowther amplifier	£60	0	0
3	Tannoy amplifier	£63	0	0

LOUDSPEAKER UNITS

-4	Tannoy 15in	£35	12	0
-4	Tannoy 12in.	£29	5	0
9	Wharfedale Super 12 C/SAL	£17	10	0
12	Wharfedale Golden CSB	£8	6	7
5	Wharfedale W10 CSB	£12	16	8
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12	Wharfedale Super 5 C/SAL	£6	13	3
7	Wharfedale Super 3 C/SAL	£6	13	3

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This tener is normality supplied with four controls—Tuning, Volume, Tone and the Wavelength Switch (Tone and Volume operate as both Radio and Gram.)—but if your Amplitier aircaidy has the Tone and Volume Controls we can omit both. When ordering plause state what is required. $0 \text{ Overal thus is dimensions are 12in. x 8jun. x 8jun. x 6lm. including the full wision dial. Size 8jin. x 4jin. Price completely assembled and landfug built-in power supply £10(10).$ H P. Frans. Deposit £21(10). & Smothas 11 29(2). Price completely assembled excluding Power Supply £2. Carriage and insurance 7/6 evers. (1) of Scout herm is 16 ever.

COMMERCIAL T/V CONVERTERS

Completely sile-contained Units designed to opence with commercially made Television. Receivers to chable the Immediate re-epton of both the B.R.C. (Band I) and the Commer-cial transmission of the Immediate re-epton of both the B.R.C. (Band I) and the Commer-cial transmission of the Immediate re-epton of both the B.R.C. (Band I) and the Commer-cial transmission of the Immediate re-epton of both the B.R.C. (Band I) and the Commer-cial transmission of the Immediate re-epton of both the B.R.C. (Band I) and the Commer-dential I immediate re-epton of both the B.R.C. (Band I) and III A extraits to the sockets provided on the Converter, which is then connected to your TiV Receiver Actual Socket. The Controls on the Unit are:--(a) Station selector Switch which Immal stative selects either Transmitting Stations. (D) An On-Off Switch which also averiable the TiV Receiver on orof. (c) Band III Station Truinic Control. THESE CONVERTINA-ARE AVAILABLE:--I. THE AERIALITE MODEL TA3. Contained In a brown reckied inished case set Splits. Ong X410. high X410. hege. Price SplitO(-, 2) THE DULCI TIV CONVERTERS. In polished walnut case size SplitO... 20 (1) high X410. Ages (or direct incorporation litbe an existing T/V receiver. In this instance it is essential that, when orienting, we are alvised of either the I.R. frequency or the make and model number of the T/V receiver. Price §8.







NOVEMBER, 1955

16		WIRELESS WOR
	RECTIFIERS H.T. Type H.W.	R.S.C.
LT. Types 2/6 v. $\frac{1}{2}$ a.h.w	120 v. 40 mA 3/11 250 v. 50 mA 5/9 250 v. 80 mA 7/9 250 v. 150 mA. 9/9	FULLY GUARAN MAINS TRANSFORMERS Primaries 200-230-250 v. 50 c/s. FULLY 8HR OUDED UPRIGE 250-0-250 v. 60 mA. 6.3 v. 2 a., Midget type, 2 <u>1</u> -3-3in. 350-0-350 v. 70 mA., 6.3 v. 2 a., 250-0-250 v. 100 mA., 6.3 v. 4 v.
	75 ohms ‡in., 7d. yard. d. yard.	0-4-5 v. 3 a. 250-0-250 v. 100 mA., 6.3 v. 4 a., 250-0-250 v. 100 mA., 6.3 v. 6 a. for R1355 conversion
30, 35, 50, 100, 120, 150 400, 470, 500, 1,000 p	NSERS. 5, 10, 15, 20, 25, 180, 200, 230, 300, 330, 161, $(.001\mu F)$, .002 mfd-ach, 3/9 dozen one type.	for R1355 conversion 300-0-300 v. 100 mA., 6.3 v. 4 a., 300-0-300 v. 100 mA., 6.3 v. 4 v. 0-4-5 v. 3 a.
DIAL BULBS, M.E.S., 6.5 v. 0.3 a., 6'9 doz. ; 4		350-0-350 v. 100 mA., 6.3 v. 4 a., 350-0-350 v. 100 mA., 6.3 v. 4 v 0-4-5 v. 3 a. 350-0-350 v. 150 mA., 6.3 v. 4 a.,
Tubular Types	tent production). ex-Govt. Can Types	350-0-350 v. 150 mA., 6.3 v. 2 a., 5 v. 3 a. 425-0-425 v. 200 mA., 6.3 v. 4 a. 4 a., c.t., 5 v. 3 a., suitable
$8\mu F$ 450 v 1/9 8 mfd. 500 v 2/6 16 μF 350 v 2/3 16 μF 450 v 2/9	16 mfd. 350 v. 1/11 16μF. 450 v. 2/9 32μF 350 v. 2/11	Amplifier, etc
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	32 mfd. 450 v 4/9 64 mfd. 450 v 3/11	5 v. 3 a. TOP SHROUDED DROP-TH 250-0-250 v. 70 mA., 6.3 v. 2.5 a 260-0-260 v. 70 mA., 6.3 v. 2 a., 350-0-350 v. 80 mA., 6.3 v. 2 a.,
$25\mu F 25 v 1/3$ $50\mu F 12 v 1/3$ $50\mu F 50 v 1/11$	100 mfd. 450 v 4/9 8-8μF 450 v 2/11 8-16μF 450 v 2/11	250-0-250 v. 100 mA., 6.3 v. 4 a., 300-0-300 v. 100 mA., 6.3 v4 v 0-4-5 v. 3 a.
8 mfd. 350 v. 1/3	16-16μF 450 v 3/11 16-32μF 350 v 4/9 32-32μF 350 v 4/9	350-0-350 v. 100 mA., 6.3 v. 4 a 3 a. 350-0-350 v. 100 mA., 6.3 v. 4 v 0-4-5 v. 3 a.
8 mfd. 600 v 2/11 16 mfd. 500 v 3/9 Many others in stock.	32-32μF 450 v 5/9	350-0-350 v. 150 mA., 6.3 v. 2 a., 5 v. 3 a. 350-0-350 v. 150 mA., 6.3 v. 4 a., E.H.T. TRANSFORMERS, 2,50
VOLUME CONTROLS values, less switch, 2/9;		2-0-2 v. 1.1 a., 2-0-2 v. 1.1 a., VCR517
WIRE WOUND POTS ohms, 5K, 20K, 100K spindles), 2/9. 220 ohm Preset type, 1/9 cach.	(medium length	THE SKY FOUR T.R.F
VIBRATORS. Wearite synchronous, 6/9. Oak 2 nous, 7/9.	v. 7 pin, synchro-	
.25 mfd., 4,000 v. Blocks .5 mfd., 2,500 v. Blocks .5 mfd., 3,500 v. Cans .1 mfd. plus 1 mfd. 8. (common negative isol 1.5 mfd., 4,000 v. Blocks	4/9 	A design of a 3-valve 200-257 receiver with selenium rectifier, either of cabinets illustrated valves 6K7, SP61, 6F6G, and is for simplicity in wring. Sensi
EX GOVT. METAL B) CONDENSERS 2 mfd. 800 v 1/9 4 mfd. 500 v 2/9 4 mfd. 1,000 v 4/3 4 mfd. 1,500 v 4/3	6-6 mfd. 450 v 5/9 8 mfd. 500 v 5/9 8-8 mfd. 500 v 5/11 15 mfd. 500 v 7/9	Is well up to standard. Poir diagrams, instructions and pair receiver can be built for a may including cabinet. Available in bakelite, or veneered walnut.
4 muu. 400 v. prus z mid	. 250 v	P.M. SPEAKERS. All 2-3 ohms 8in. Rola Heavy Magnet, 19/9. 10in. Rola 27/9
EX GOV. UNITS. type cartons 29/6. Transmit complete with all valve:		R.S.C. BATTERY CHARGER input 200-250 v. 50 c/s. To cha lator a To cha
Yaxley switch, 1 pointer various plugs and soci	ith 1 six-position 3-wafer knob, 2 S.P.S.T. switches, kets. Only 1/6. 2-3 ohms, 8in. R.A. field.	battery To chas battery ABOV
MANUFACTU	A. field, 1,000 ohms, 23/9. RERS SURPLUS	OF GH LOUV CASE,
Fully shrouded uprigh	ORMERS . Primary 200-230-250 v. m.a. 6.3 v. 3 a. 5 v. 3 a. 00-0-250 v. 70 mA., 6.3 v.	FORMER, FULL WAVE MET FUSES, FUSE-HOLDERS
HEAVY DUTY BATT For normal 200/250 v charge 12 v. battery.	ERY CHARGER A.C. mains input. To Variable charge rate of up er and Fuses. Guaranteed	Variable charge rate of
OIL FILL COND Bryce 11-7 mfd. 500 surplus, only 5/9 each	ED BLOCK ENSERS v. New unused Govt.	up to 4 AMPS. Fused, and with ammeter. Well ventilated metal case with attractive crackle finish. Guarankeed for 12 months, 69/6. Carriage 2/6.
KIT with louvred crac		BATTERY CHARGER KIT. Bridge Rectifier 6/12 v. 5 a. M 250 v. input, 0-9-15 v. 6 a. ou charge rate. Rheostat with know

INS TRANSFORMERS maries 200-230-250 v. 50 c/s. LLY SHROUDED UPRIGHT MOUNTING +0-250 v. 60 mA. 6.3 v. 2 a., 5 v. 2 a., 17/6 +0-350 v. 70 mA., 6.3 v. 2 a., 5 v. 2 a., 19/9 +0-250 v. 70 mA., 6.3 v. 4 v. 4 a., ct., +4-5 v. 3 a. -0-250 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a 23/9 +0-250 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a 23/9 +0-300 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a 23/9 +0-300 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a 23/9 +0-300 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a 23/9 +0-300 v. 100 mA., 6.3 v. 4 v. 4 a., ct., -0-30 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a 23/9 +0-350 v. 100 mA., 6.3 v. 4 a., 5 v. 3 a 23/9 +0-350 v. 100 mA., 6.3 v. 4 v. 4 a., ct., -4-5 v. 3 a. -26/9	LEAVED AND IMPREGNATED FILAMENT TRANSFORMERS Primaries 200-250 v. 50 c/s. 0.4-6.3 v. 2. 6.3 v. 1.5 a 5/9 0-2-4-5-6.3 v. 6.3 v. 3 a 8/11 6.3 v. 6 a. 12 v. 1a 7/9 12 v. 3 a. or 6.3 v. 2 a 7/6 1.5 a CHARGER TRANSFORMERS All with 200-230-250 v. 50 c/s. Primarie 12 a. 11/9: 0-9-15 v. 5 a., 0-3.5-9-17 v. 0-9-15 v. 5 a., 19/9: 0-9-15 v. 6 a., 23/9 ELIMINATOR TRANSFORMERS Primaries 200-250 v. 50 c/s. 120 v. 40 130 v. 50 mA., 6-0-5 v. 1 a. 90 v. 15 mA., 6-0-5 v. 1 a.
$\begin{array}{c} \text{-0-350 v. 150 mA., 6.3 v. 4 a., 5 v. 3 a 33/9} \\ \text{-0-350 v. 150 mA., 6.3 v. 2 a., 6.3 v. 2 a., 5 v. 3 a.} \\ \text{-0-250 v. 200 mA., 6.3 v. 4 a., c.t., 6.3 v. 4 a., c.t., 5 v. 3 a.} \\ \text{-0-250 v. 250 mA., 6.3 v. 4 a., c.t., 6 v.} \\ -0-250 v. 250 mA., 6.3 v. 6 a., 6.3 v. 6 a., 6 v. 6 v. 7 0 mA., 6 v. 2 a., 5 v. 2 a., 16 v. 9 v0 v. 7 0 mA., 6 v. 2 a., 5 v. 2 a., 16 v. 9 v0 v. 100 mA., 6 v. 4 v. 4 v. 4 a., ct., 0 v. 4 v. 4 a., ct., 0 v. 4 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 a., ct., 5 v. 3 a., 10 mA., 6 v. 4 v. 4 v. 4 v. 4 v., 4 v. 4 v. 4 v. $	$\label{eq:constraints} \hline \begin{array}{ c c c c c c c c c c c c c c c c c c c$
0-4-5 v, 3 a	SMOOTHING CHOKES 250 mA., 5 H., 100 ohms 150 mA., 7-10 H., 250 ohms 100 mA., 10 H., 150 ohms potted 100 mA., 10 H., 150 ohms s. 80 mA., 10 H., 350 ohms 60 mA., 10 H., 400 ohms
E SKY FOUR T.R.F. RECEIVER	EX GOVT. MAINS TRANS All 230 v. 50 c/s. input. 8.8 v. 4 a. 9/9 48 v. 1 a. 978 v. 200 mA. 7/9 24 v. 1 a. 750 v. 4 times (high ins.) . 6.3 v. 10 a. 120-0-120 v. 40 mA. 300-0-300 v. 150 mA. 4 v. 3 a. 250-0-250 v. 60 mA. 6 3. v. 2 a., 2 a. Potted

valve 200-250 v. A.C. Mains jum rectifier. For inclusion in illustrated above. It employs of 6G, and is specially designed viring. Sensitivity and quality undard. Point-to-point wiring ons and parts list, 1/9. This ilt for a maximum of t4/19/6 Available in brown or cream ed walnut. ed walnut.

All 2-3 ohms. 61in. R.A. 15/9. agnet, 19/9. 10in. R.A. 25/9.



CHARGER KITS. For mains 0 c/s. To charge 6 v. accumu-lator at 2 amps., 25/9. To charge 6 v. or 12 v. battery at 2 a., 31/6. To charge 6 v. or 12 v. battery at 4 a., 49/9. ABOVE KITS CONSIST OF GREEN CRACKLE OF GREEN CRACKLE LOUVRED STEEL CASE, MAINS TRANS-

WAVE METAL RECTIFIER, OLDERS AND CIRCUIT. ed and tested for 6/9 extra.



BATTERY CHARGER KIT. Consisting F.W. Bridge Rectifier 6/12 v. 5 a. Mains Trans. 200-250 v. input, 0-9-15 v. 6 a. output and variable charge rate. Rheostat with knob 45/9.

FILAMENT TRANSFORMERS	
Primaries 200-250 v. 50 c/s. 0-4-6.3 v. 2 a	7/9
6.3 v. 1.5 a 5/9 0-2-4-5-6.3 v. 4 a. 6.3 v. 3 a 8/11 6.3 v. 6 a.	16/9
6.3 v. 3 a, 8/11 6.3 v. 6 a	17/6
12 v. 1 a 7/9 12 v. 3 a. or 24 v. 6.3 v. 2 a 7/6 1.5 a	17/6
	1//0
CHARGER TRANSFORMERS	
All with 200-230-250 v. 50 c/s. Primaries: 0-5 12 a., 11/9; 0-9-15 v. 3 a., 0-3.5-9-17 v. 4 a.,	
0-9-15 v. 5 a., 19/9; 0-9-15 v. 6 a., 23/9.	10/3-
	-
ELIMINATOR TRANSFORMERS	7/11
Primaries 200-250 v. 50 c/s. 120 v. 40 mA.	14/0
130 v. 50 mA., 6.3 v. 3 a	14/9
120 v. 40 mA., 5-0-5 v. 1 a.	9/11
OUTPUT TRANSFORMERS	
Midget Battery Pentode 66:1 for 3S4. etc	3/6
Small Pentode, 5,000 Ω to 3 Ω	3/9
Standard Pentode, 5,000 Ω to 3 Ω	
Standard Pentode, 8,000 to 3Ω	4/9
Battery Pentode, 10,000 ohms to 3 ohms Multi-ratio 40 mA. 30:1, 45:1, 60:1, 90:1,	4/9
	5/6
Push-Pull 8 Watts 6V6 to 3 ohms	8/9
Push-Pull 10-12 Watts 6V6 to 3Ω to 15Ω ,	
sectionally wound	16/9
	16/9
3-5-8 or 15Ω Push-Pull 15-18 Watt, sectionally wound,	10/9
6L6. KT66, etc., to 3 or 15 ohms.	21/9
Push-Pull 20 Watt high-quality sectionally	
wound, 6L6, KT66, etc., to 3 or 15Ω	47/9
SMOOTHING CHOKES	
250 mA., 5 H., 100 ohms	11/9
150 mA., 7-10 H., 250 ohms	11/9
100 mA., 10 H., 150 ohms potted 100 mA., 10 H., 200 ohms	9/9
80 mA., 10 H., 350 ohms	819
60 mA 10 H 400 ohms	2.0
	. A .
EX GOVT. MAINS TRANSFORM	MERS

TRANSFORMERS

EX GOVT. MAINS TRANSFORMERS
All 230 v. 50 c/s. input. 8.8 v. 4a. 9/9 48 v. 1 a
Carriage on following types 5/- extra. 0-11-22 v. 30 a
EX GOVT. AUTO TRANSFORMERS 15-10-5-0-195-215-235 v. 500 Watts 27/9 Double wound 10-0-200-240 v. to 10-0-275- 295-315 v. Series connection will make suitable for 110 v. to 230-250 v. or reverse, 1,000 watts
EX GOVT. SMOOTHING CHOKES 250 mA., 10 H., 50 ohms 14/9 250 mA., 3 H., 50 ohms 8/9 150 mA., 10 H., 50 ohms 10/11 150 mA., 10 H., 150 ohms, Tropicalised 6/9 100 mA., 10 H., 100 ohms, Tropicalised 3/11 50 mA., 50 H., 1000 ohms, Tropicalised 3/11 50 mA., 50 H., 1000 ohms, 200 ohms,
SPECIAL OFFERS Filament Transformers. Primaries 230-250 v. 50 c/cs. 2 v. 2 a. 4/9. 4 v. 2 a. 4/9. 6.3 v. 1 a. 4/9
$ \begin{array}{c} \textbf{CHASSIS} \\ \textbf{16} & \textbf{s.w.g. undrilled aluminium amplifier type} \\ (4-sided). \\ \textbf{14in. \times 10in. \times 3in. 7/11} \\ \textbf{16in. \times 10in. \times 3in. 7/11} \\ \textbf{16in. \times 10in. \times 3in. 8/3} \\ \textbf{18} & \textbf{s.w.g. aluminium reciver type.} \\ \textbf{12in. \times 8in. \times 2 \pm in. 7/6} \\ \textbf{20in. \times 8in. \times 2 \pm in. 8/1} \\ \textbf{16} & \textbf{s.w.g. aluminium reciver type.} \\ \textbf{10} & \textbf{10} & \textbf{10} & \textbf{10} \\ \textbf{10} & \textbf{10} & \textbf{10} & \textbf{10} \\ \textbf{10} & \textbf{10} & \textbf{10} & \textbf{10} \\ \textbf{10} & \textbf{10} & \textbf{10} & \textbf{11} \\ \textbf{10} & \textbf{10} & \textbf{10} & \textbf{10} \\ \textbf{10} & \textbf{10} & \textbf{11} \\ \textbf{10} & \textbf{10} & \textbf{10} \\ \textbf{10} & \textbf{10} & \textbf{10} \\ \textbf{10} & \textbf{10} & \textbf{11} \\ \textbf{10} & \textbf{10} & \textbf{11} \\ \textbf{10} & \textbf{10} & \textbf{11} \\ \textbf{10} & \textbf{10} & \textbf{10} \\ \textbf{10} & \textbf{10} \\ \textbf{10} & \textbf{10} \\ \textbf{10} & \textbf{10} & \textbf{10} \\ \textbf{10} & \textbf{10} \\ \textbf{10} & \textbf{10} & \textbf{10} \\ \textbf{10} & \textbf$

R.S.C. HIGH FIDELITY 25 watt AMPLIFIER

A NEW DESIGN FOR 1955 HIGH GAIN "PUSH PULL OUT-PUT". BUILT-IN PRE-AMP. TONE CONTROL STAGES. INCLUDES valves, sectionally wound output transformer, block paper reservoir condenser, and reliable small components. AN INPUT OF ONLY 20 millivolts IS REQUIRED FOR FULL OUTPUT. THIS MEANS THAT ANY TYPE OF MICROPHONE OR PICK-UP IS SUITABLE. Two separate inputs controlled by separate volume controls allow simultaneous use of "Mike" and Gram., or Tape and Radio, etc., etc.

of 24 D.B. Frequency response \pm 3 D.B. 30-20,000 c/s.

COLLARO HIGH FIDELITY 3.SPEED MIXER AUTO-CHANGERS, TYPE RC/54, Latest model fitted with Studio "O" Turnover Pick-up Head, Very imited number, Brand New, Guaranteed, Only 10 gns, carriage 7/6. Or Deposit 4 Gns. 7 monthly payments 1 Gn.

COLLARO HIGH FIDELITY MAGNETIC PICK UPS. Low impedance with matching trans, brand new, boxed at fraction of normal price. Only 35/nd

A PUSH PULL 3-4 WATT HIGH GAIN ASSEMBLED AMPLIFIER FOR £3/19/6. For mains input 200-250 v. 50 c/s. Amplifier can be used with any type of feeder unit or pick-up. This is not A.C./D.C. with "live" chassis but A.C. only with 400-0-400 v. Trans. Output is for 2-3 ohm speaker. Supplied ready for use. £3/19/6. Leaflet 6d.

H.M.V. LONG PLAYING RECORD TURN-TABLE COMPLETE WITH CRYSTAL PICK-UP (SAPPHIRE STYLUS). Speed 334 r.p.m. BRAND NEW, CARTONED. Only £3/19/6 (approx. half price). Cart. 5/- (for 200-250 v. A.C. Mains).

A7 3-4 WATT QUALITY AMPLIFIER R.S.C R.S.C. A7 3-4 WATT QUALITY AMPLIFIER A highly sensitive 4-valve amplifier using negative feedback and having an excellent frequency response. Pre-amplifier and Tone Control stages are incorpor-ated with separate Bass and Trebie controls giving full tone compensation for Long Playing records. Sultable for any kind of plck-up including latest high fidelity types. M.T. of 250 v. 20 mA. and L.T. 6.3 v. Ia. avail-able for supply of Radio Feeder Unit, etc. ONLY 40 millivoits input required for full output. Fully isolated chassis with baseplate. For A.C. mains 200-250 v. 50 cycles. Output for 2-3 ohm speaker. Complete kit of parts with point-to-point wiring diagrams and instructions. Only £3/15/-.

R.S.C. 4-5 WATT HIGH GAIN AMPLIFIER TYPE A5



R.S.C. 4-5 WATT HIGH GAIN AMPLIFIER TYPE AS A highly sensitive 4-valve quality amplifier for the home, small club, etc. Our 50 millivolics for un 5 sensitive for use with the latest find-fidelity pick-up heads, in addition to all other types of pick-ups and practically all mikes. Separate Bass and Trebic controls are provided. These give full long playing record equal-sation. Hum level is negligible being 71 D.B. down. 15 D.B. of negative feedback is used. H.T. of 300 v. 25 mA. and L.T. of 6.3 v. 1.5 a. Is available for the supply of a Radio Feeder Unit, or Tape Deck pre-amplifier. For A.C. mains input of 200-230-250 v. 50 c/s, Output for 2-3 ohm speaker. Chassis is not alive. Kit is complete in every detail and includes fully punched chassis (with baseplate), with green crackle finish, and point-opoint witing dlagrams and instructions. Exceptional value at only £4/15/-, or assembled ready for use 25/- extra, plus 3/6 carriage.

MICROPHONES. High fidelity crystal types. Acos 33-1 hand or desk type 50/-. Piezzo with heavy floor base and telescopic stem £6/19/6.

GOLDRING MAGNETIC PICK-UPS. Due to a fortunate purchase we can offer these popular high impedance pick-ups. Brand new, boxed, at only 23/9

ų

Hum level 66 D.B. down. Certified total harmonic distortion of only 0.35% measured at 10 watts. Comparable with the very best designs. SUITABLE FOR SMALL HOMES OR LARGE HALLS, CLUBS. GARDEN PARTIES. DANCE HALLS, etc., etc. For ELEC-TRONIC ORGAN OR GUITAR. For STANDARD OR LONG PLAYING **RECORDS.** Size 12 \times 10 \times 9 in. For mains A.C. 200-250 v. 50 c/s. Power consumption 175 watts. Outputs for 3 and 15 ohm speakers. The kit is com-plete in every detail. Chassis is fully punched. Easy to follow point-to-point

Individual controls for Bass and Treble "lift" and "cut". Six negative feedback loops giving total 121 D B 20 20 000 c/c Or commission of the sense Or assembled ready for use 50/- extra. GNS. U H.P. Terms on assembled units. Deposit 26/- and 12 monthly payments of £1. Plus carr. 10/-.

Terms to include cover, microphones, speakers, etc., on request. Cover as illustrated if required, price 17/6 extra.



COLLARO 3-SPEED AUTO-CHANGERS. New guaranteed units. Flitted 'Plug in' crystal pick-up heads for standard or long playing records. Very linuited number available at approx. half price. For A.C. mains 200-250 v. 50 c/cs. Only 27/19/6. Carr. 716

DEFIANT RECORD PLAYING TURNTABLE COMPLETE WITH MAGNETIC PICK-UP. PIck-up is high impedance type. Unit is housed in a beauti-ful walnut vencered cabinet of attractive design. For all standard records (78. r.p.m.), Limited number. Brand new, cartoned, £5/19/6. Carr. 7/6.

BAKER SELHURST QUALITY SPEAKERS. 12in. 15 ohm 15 watt Stalwart £5/15/-. 12in. 15 ohm 20 watt Standard £7.

R.S.C. MASTER INTERCOMM. UNIT, with provision for up to 4 "Listen-Talk Back Units" indi-vidually switched. A high ratin amplifier enables speech and other sounds emanating from the rooms containing remote control units to be heard at the master control. Supplied with walnut veneered wood or brown backlite cabinet. Mains input is 200-250 v. 50 cis. H.T. line 300 v. CHASSIS IS NOT "ALIVE." Ideal for use as "Baby Alarn." Sound amplification 4 watts. Price only 7 gns., carr. 5/-, "Listen-Talk Back Unit' in backlite or walnut veneered cabinet can be supplied at 35/- each.



ALL DRY RECEIVER BATTERY ELIMINATOR KIT

All parts for an "All Dry" Battery Elimina-tor, complete with case. Completely replaces 1.4 v. and 90 v. batterles where normal mains

1.4 v. and 90 v. batterles where normal mains supply of 200-250 v. 50 c/s. is available. Price with circuit, 38/9 Or ready for use, 45/6 L.T. loads of 125 mA. to 250 mA., thereby covering latest low consumption types.

BATTERY SET CONVERTER KIT. All parts for DATERX SET CONVERTER KIT. All parts for converting any type of battery receiver to all mains. A.C. 200-250 v. 50 c/s. Kit will supply fully smoothed H.T. of 120 v. 90 v. or 60 v. at up to 40 mA. and fully smoothed L.T. of 2 v. at 0.4 a. to 1 a. Price complete with circuit and instructions only 48/9. Supplied ready for use for 8/9 extra.



Arre safety factors in every component A.C. and H.T. duss, bunched chassis with baseplate, screened hopoint withing diagrams. Everything supplied to lay supplied to a series of the series of the series of the problem of the series of the series of the series and the series of the series of the series of the series and the series of the series of the series of the second series of the second series of the series

COMPLETE Kit of Parts 7 GNS, (carriage 7/6).

Supplied, assembled and tested for 45/- extra. Cover as for A4 amplifier 17/6 extra if required.

H.P. TERMS on assembled units. Deposit 23/6 and 9 monthly payments 21/-.

FOUR-STAGE RADIO FEEDER UNIT

FOUR-STAGE RADIO FEEDER UNIT Design of a HIGH FIDELITY L. and M. wave T.R.F. Unit with self-contained heater supply and thorough H.T. decoupling. Only 250-400 v. 15-20 mA. H.T. required from main ampilier. Three vaives and Low Distortion Germanium Diode Detector. Flat topped response characteristic. Loaded H.F. colis. Two variable Mu controlled H.F. stages. 3 gang con-denser tuning. Cathode follower output stage. Switch position for Gram. and Gram. input and output sockets. Performance comparable with the best in Feeder Units. For A.C. mains 200-230-250 v. opera-tion. Size 116-716. Illustration, full set of easy-to-follow wirling diagrams and instructions and indi-vidually priced parts list 21/6. This unit can be built for only ±3/15/r., including Dial and Drive Knobs and every item required.

W.B. "STENTORIAN" HIGH FIDELITY P.M. SPEAKERS, HF1012, 10 wats, 15 ohm (or 3 ohm) speech coil, Where a really good quality speaker at a low price is required we highly recommend this unit with an amazing performance. £4/2/9.



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NOVEMBER, 1955

HIRE PURCHASE

only, plus 3/6 packing and postage. RECEIVER TYPE 25/73. (The receive's section of TR 1196). Supplied complete which the data for contribution to 3-wave-which the data for contribution to 3-wave-which the data for contribution to 3-wave-which the data for contribution to 3-wave-sector 2-EF39. LEF36. FK32 into EBC33, also standard 1.F.T.'s 465 Ke's. Price 27/6 plus 2./6 P. and P. TRI196 TRANSMITTER PORTION. We can also supply the transmitter por-tion of the above receiver incorporating valves, EL32. EF50. CV501. Type 600 relay transformer, colls. switches, etc. Limited quantity at 12/6 only, plus 2/6 P. and P. CRYSTAL MIKE INSERTS. Brand

CRYSTAL MIKE INSERTS. Brand new by Cosmocord. Price 7/6 cach only. Post free.

Post free. 22 SET POWER UNITS No. 4MK1 ZA10478—Complete with 4 metal recti-fiers each 250 v. 60 mA. 2-12 v. 4 pln Mallory Vibrators, transformers, con-densers, resistors, signal I amp, indi-cator, etc., etc., in good condition. Com-plete in metal box size 10 jin. X 6in. X 81n, Weight 191b., 27/6, plus 5/- P. & P.



F.M. !! (Frequency Modulation)



Introducing our latest F.M. Tuner. Of advanced design, employing new technique. *Tuned resonator R.F. Stage. *Ultra-stable co-axial oscillator. *High sensitivity. *Gorler I.F.T's and discriminator. This tuner is completely stable with no warm-up drift. Easy to construct and align. The ready-drilled chasts not only includes dial and drive assembly com-plete, with tuning condenser, but volume control ready mounted. Attract-ively mished in bronze, black and gold, dial ready calibrated in megacycles. From panel measures 8 Hin.×Sin.. dial Sin. X lave line-up is 4-6AM6 er equivalent. Illustrated comprehensive instruction booklet with individually-priced component iis! 1/6 post free. Or, the kit complete right down to the last tut and bolt 56/19/6, plus 2/6 P. & P.

the kit complete right down to the last nut and bolt $$\delta(1)\delta(b)$, bius 2/6 P. & P. U.S.A. PACKARD-BELL PRE-AMPLIFIER. Incorporations valves GLTGT, 28D L. relations valves GLTGT, 28D L. relations with mitructers booklet, 12/6 only. MAINS TRANSFORMER BAR. Galves TRANSFORMER BAR. Galves TRANSFORMER BAR. Galves J. Limited quantities. Manu-facturers' surplus 350-0-350, 120 mA, 6.3 v, 3 a., 5 v, 2 a. Half shrouded. drop-through. 14/6 only. plus 1/6 P. & P. 110/210/240 v. Input. 350-0-350, 120 mA, 6.3 v, 6 A., 6.3 v. 15 A., 5 v. 3 A. tropicalised drop-through type, 21/-only, plus 2/6 P. & P. 110/210/240 v. In-put, 250-0-250 120 mA., 6.3 v. 4 A., 5 v. A. Upright mounting, 21/- plus 2/-P. & P. 230 v. Input, 300-0-300 80 mA. G. 3 v. 3 A., 4 v. 2 A. Tropicalised drop-through type, 9/6 only, plus 1/6 P. & P. Input 110/230 v. Auto Ioad 230 v. 750 mA, 350-0-350 130 mA. Tapped filament 9. Carlo, 100 m/a. 6.3 v. 3 A., 5 v. 2 A. Or 270, 100 m/a. 6.3 v. 3 A., 5 v. 2 A., 200/ 250 v. Input universal mounting 16/6. Just 16/6 P. & P.

LT. TRANSFORMER. Manufacturer's surplus. Input 180/200/220/250 v. Out-put 6 v. 2 a. and 6 v. 6 a., 15/- only, plus 1/6 P. G P.

Surplus. Input 180/2007230/250 v. Out-put 6 v. 2 a. and 6 v. 6 a., 15/- only, plus 1/6 P. G P. TELESCOPIC AERIAL MAST. Ex-R.A.F. dinshy transmitter mast. Total length when extended, 17/ft. Collapses into two sections each approx. 24/n. Complete with dies and lashings, light-weight duralumin construction, dia, at thickest point, 13/n. approx, tapering to 3/n. New condition, 32/6, plus 2/-post and packing. EX-W.D. CATHODE RAY TUBES. Guaranteed full picture. VCR97 at 40/-VCR517C at 35/-. Also VCR139A-dical for oscilloscope 24/n. screen at 35/-. We also have VCR97 with slight nexting purposes, etc., at 15/- only, Al these tubes are brand new. In original packing, and tested before despatch. Please add 2/6 packing and Carriage for any of the above tubes. TRANSISTORS. Multard Type OC71 available from stock, 40/- post free. AMERICAN INDICATOR UNIT TYPE BC929A. Brand new incorpora-ing 3/n. tube 3BPI, with mu-metal shield 2-65N70T, 2-6H6GT. 6X5G, 2X2, GG6G, 9 potentiometers 24 v. actial switch motor, transformer and a host of small components. The whole unit which measures only 83in.X84in.X 134in. Is brand new, enclosed th black crackie box, and can be supplied at 65/-, plus 5/- P. & P.

We are pleased to announce our complete Kit is on the "Denco" F.M. Feeder Unit. This unit provides an A.F. output suitable for feeding into the audio section of a standard broadcast receiver where triody pentode output are available. Within an average of 30 miles from a V.H.F. transmitter one I.F. stage should be adequate, but our complete Kit supplied includes all components and values for an extra I.F. stage fill Constructional details, theoretical circuit and point-to-point witing diaram can be supplied for 1/6 post free, or the complete Kit right of the 2/6 mole and both at onto the supplied includes all be extended to the supplied of the state of the supplied of the state of the supplied of the state of the state of the supplied of the state of the state of the state of the state of the supplied of the state of the state of the state of the state of the supplied of the state of the st



booklet by Data Publications, price 2/- post free, With each booklet is enclosed our individually priced parts list. The construction and alignment of this tuner are no more difficult than a normal medium wave tuner. It is highly sensitive and free from drift. Incorporates 4 valves type 6AM6 and 2 specially graded G.E.C. Crystals. The kit supplied includes drilled chassis with tuning condensers, scale calibrated in tuning condenser, scale calibrated in megacycles, and attractive bronze stove enamelled front plate aiready nounted (as illustrated) front plate

size 8in. × in., chassis size 7in. ×44in. ×14in. size bin. \times 1n., Chassis size /in. \times 4jin. \times 1jin. N.B. The standard model is at present operating satisfactorily up to 80 mlles from Wrotham. Our price for the complete standard kit is $\pounds/15/$ -only ! Plus 2/6 p. & p. Fringe area model including extra valve, coil, etc. (results could be expected up to 150 miles from Wrotham), is $\pounds/15/$ -, plus 2/6 p. & p. The Standard Model Tuner can be supplied ready built, aligned, tested and manufactured by the Jason Motor and Electronic Company at a price of $\pounds15/17/$ -, purchase tax paid.

N.B.—THESE TUNERS ARE BEING DEMONSTRATED AT 18 TOTTEN-HAM COURT ROAD.

F.M. AERIALS. Indoor two-element type by Lumex. Brand new 11/6 each only, plus 2/- postage and packing. Other types available.

F.M. POWER PACK KIT.—We can now supply complete kit for power pack suitable for either of the above F.M. tuners or any other similar type. Price for the complete kit is 37/6 for ready assembled unit. This pack is extremely small incorporating valve rectifier type 6X4 and built on chassis size only 6in. X4in. X1ih. Optional extra for power pack. Builgh Octal Plug 2/3.

			MET	TERS	
	F.S.D.	Size	Туре	Fitting	Price
and a second	50 mlcroamp 50 mlcroamp 100 mlcroamp 500 mlcroamp 1 mA. 1 mA. 1 mA. 5 mA. 50 mA. 150 mA. 200 mA. 200 mA. 200 mA. 200 mA. 1 amp. 1 amp.	D.C. 21n. D.C. 24in. D.C. 24in. D.C. 2in. D.C. 2in. D.C. 2in. D.C. 2in. D.C. 2in. D.C. 2in. D.C. 2in. D.C. 2in. D.C. 2in. D.C. 2in. R.F. 2in. A.C. 4in. R.F. 24in.	Type M.C. M.C. M.C. M.C. M.C. M.C. M.C. M.C	Fitting R.P. F.R. Posk Type F.Sq. R.P. R.P.	50/ 65/- 45/- 18/6 17/6 22/6 227/6 30/- 7/6 8/6 7/6 8/6 10/- 6/6 10/- 7/6
	R.P. = Round I F. Sq. = Flush	D.C. 211n. A.C. 211n. A. 0-1 mA. 2 Projection. M Square, F.R.	.C. = Mo = Flush	F. Sq. F. Sq. Thermo F.R. R.P. (with shund) F.R. F.R. F.R. F.R. F.R. from equipment but perfect. 22/6 volng Coll. Thermo \Rightarrow Thermo-con Round. M.I. \Rightarrow Moving Iron. C., at 8/6, also 5 mA. by G.E.C. a	upled.

This kit has been based on the booklet by Data Publications, price



Price only 30/- each. ALL these units Post Free 1 I.F. TRANSFORMERS SPECIAL OFFER. All Iron-cored 465 Kc/s. By Invicta — Cylindrical 24in x 18in. (dnm, 8/6 pr. Also our own spin-ultra midget size 13in x 13i 16in x 13/ 16in. Only 9/9 pr Bw Wearlet, Tyre 501 and 502 12/6 pr. Tyre 531 and 552 12/6 pr. M800 12/6 pair. COLL PACKS. Manufacturers' Surplus. Minature size, only 21in x 28in. X 14in. deep. Iron-cored. For L.M. and S.W. with gram, position. Switch has 2-inch spindle. Absolutely brand new, com-piete with circuit. Price only 25/-, plus 1/6 pr. & P. A snip 1 HEADPHONES. Brand new, c.Gort. by S.G. Brown. Type CLR. Low resistance. 71/6 ner pair. No. 38 TRANSMITTER/RECEIVER WALKIE-TALKIE. Range approx. 5 miles. Coverage 7.4-9 Mc/s. The set only complete with valves at 30/-, in very zood condition.

Please add postage under £1, C.O.D. or Cash with order. C.O.D. charge extra—open 9 a.m. to 6 p.m. Monday to Friday. Sorry, but we close at 1.0 p.m. on Saturday.

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Mains, 2007 Main

The R.E.P. ONE-VALVE BATTERY RECEIVER KIT. Simple one-valve all dry battery receiver (or head-phones, easily built in one evening. All required components including headphones, can be supplied at inclu-sive cost of 42/- plus 2/- p. and p. Operated by Ever-Ready B114 type battery available at 7/9. Full assembly details available separately at 9d. plus 3d. post.

T.S.L. ELECTROSTATIC LOUD-SPEAKERS ! A much-wanted need in High Fidelity reproduction! Model LSH75. Size only 3in. X 3in. X 4in. Weight 162. only. Capacity 800 pf. D.C. voltage: 300 max. Tone frequency A.C. voltage: 60 volts max. effective. Test voltage: 61 volts max. effective. These speakers have a high efficiency in the range of 5,000-20,000 cycles, and are a must for F.M. high quality recordings, and T.V. sound ! Each speaker supplied with full technical data, response curve, and wiring diagram. T.S.L. ELECTROSTATIC LOUD-SPEAKERS ! A much-wanted need



posit plus p. and p. and 12 monthly payments of 18/7.

PLAYING DESK! Two-speed 33 and 78 r.p.m. player by famous manufacturer. Complete with turn-over crystal pick-up, Already mounted on platform, ready to use, £5/19/6 only, plus 5/- P, and P.

B.S.R. MONARCH. The very latest crean 3-speed mixer Auto-changer. Complete with turn-over crystal pick-up. Complete in original manufacturer's cartons, fully guaran-teed. Price only £7/19/6. Buy now I Quantity at this price strictly limited.

TABLEGRAM CABINETS. Manu-facturer's Surpius I Handsome dark walnut finish, Size 164in.×134in.× 114in. high. Motor-board already cut for latest type B.S.R. Monarch Autu-changer. Provision at side for amplifier controls. Price 79/6, plus 5/. P. & Baffle fitted for 71n.× 41n. Elliptical speaker for which we can supply latest ROLA at 21/6. speaker for whi ROLA at 21/6.

ROLA at 21/6. RECORD PLAYER CABINETS. Specially made to house any type of single record unit. Finished in dove-grey leathcrette. Baseboard measures 14jin. x 12jin. Clearance above and below board 3in. 45/- pus 3/- P. & P. We can also supply equally attractive dove-grey cabinet to house any standard auto-changer at 69/6 plus 3/- P. & P. We carry a large selection of cabinets for all purposes. A stamp will bring illustrated cabinet leaflets.

VALVES We have perhaps the most up-to-date valve stocks in the trade. A stamp will bring complete list but the following is a selection only of brand new imported valve types, fully guaranteed. Purchase valve type Tax Paid. DAE9610/6 PY80 EADCOO

LADCOU	DAF9010/0	P Y 80	10
10/-	DF96 10/6	PY81	10
EAF42 10/-	DK92 10/6	PY82	9
EB41 7/6	DK96 10/6	PY83	11
EB91 7/6	DL96 10/6	UBC4	
EBC4110/-	or 39/6 per		
EBF8011/6	set of four.	UCH4	
ECC81 9/-	EL41 10/6	UF41	
ECC82 9/-	EL84 11/6	UL41	
ECC83 9/-	EM80 9/-	UY41	
ECC8510/-	EY51	6AQ5	8
ECH4211/6	(large) 11/-	6AT6	8
ECH8111/6	EZ40 8/6	6AU6	9
ECL8011/6	EZ80 8/6	6BA6	8
EF41 10/6	PCF80 12/6	6 BE6	9
EF80 10/6	PCF8212/6	6BW6	8
	PCC8412/6	6X4	7
	PL81 13/6	35W4	7
EF86 12/6	PL82 10/6	50B5	10
EF89 10/-	PL83 11/6	50C5	10

In addition we naturally have all usual surplus types available such as 6V6GT, etc. All in our valve price list !

N.B. All our T.R.F. Kit circuits now include specially wound Denco "Maxi-Q" coils on polystyrene formers, improved performance 1 Price remains the same

THE "ECONOMY FOUR" T.R.F. KIT A three-valve plus metal rectifier receiver. A.C. mains 200/250 v, Medlum and Long waves. We can supply all required compo-nents risht down to the last nut and bolt. Valve line-up 6K7 617 and 6V6. Chassis ready drilled—Cabinet size 12in. long by 6in. hish by 5in. deep—Choice of lyory or brown Bakelite, or wooden, walnut finish cabinet. Complete instruction booklet with practical and theoretical diagrams. Each component brand new and tested prior to packing. Our price £5/10/- complete—Remember this set is being demonstrated at our shop premises. We proudly claim that our fully illustrated Instruction booklet is the most comprehensive available for this type of receiver —Booklet available at 1/6 post free. This is allowed if kit is purchased later. Please. 2/6 packing and carriage for complete kit.



Please. 2/6 packing and carrlage for complete klt. THE R.C. GRAM REPLACEMENT CHASSIS KIT To meet the very great demand for this type of receiver, we have produced this unit. For Long, Medium and Short Waves. Valve line-up: 6K8 Frequency changer. 6K7, I.F. Amplifier, 6Q7 Ist Audio Detector and A.V.C. 6V6 Output, SX5 Full-wave reclifter, For A.C. mains 200/250 volts. 4 watts output. Exectlent guality. High sensitivity. Provision for green and gold dial for horizontal tuning. Four controls are: Tuning, L/M/S. Gram. Vol./on/off. Tone (vari-able). Chassis size: 134 in. X54in. X5 2 in. Dial size: 10in. X4 in. Assembly is simplified by the use of a 3-waveband coll pack, and pre-aligned 465 Kc/s. I.F. transformers-high-arede drop-througeh

simplified by the Use of a 3-Wayeband coil pack, and pre-aligned 455 Kc/s. LF, transformers—high-grade drop-through half-shrouded Mains Transformer, with voltage adjuster panel. This chassis can easily be assembled in one evening. Illustrated pamphiet with full assembly instructions, practical and theoretical and drive at 39/6. 3-wayeband college to 455 Kc/s. LF. Transformers, 9/6 pair. Half shrouded dray-stormlete with and college adjuster panel, the stormers, 9/6 pair. Half shrouded dray-through Mains Transformer, 22/6. The total cost of ALL litems purchased separately is nearly \$10, but we shall be pleased to supply all the required components right down to the last nut and boil at a special inclusive price of \$8/8/1, plus 2/6 packing and postage. A set of four small brown and cream engraved knobs to suil is available at 1/2 each knob. This chassis is a professional job in every respect and can be seen and heard at our premises. This chassis can also be supplied, ready assembled in very limited quantities at \$2/19/6, plus 5/- carriage and packing. ARMSTRONG E.C.48. Their very latest high quality replacement chassis

ARMSTRONG F.C.48. Their very latest high quality replacement chassis having provision for F.M. feeder unit. 8 valves, 4 wavebands. Independent bass and treble with unique thermometer visual indicator. Ready for use £23/18/-plus 5/- package and postage or £5/18/- deposit and 12 monthly payments at 33/9. Illustrated leaflet available.

DULCI RADIO/RADIOGRAM CHASSIS. All latest models including F.3 and F.3 push-pull are in stock. Cash or H.P. Ask for illustrated leaflet.

COLLARO 2010. Transcription motor with Studio Pick-up. This very popular unit can now be supplied from stock. £18/5/3 cash or 95/3 deposit, and 12 payments of 25/8.

London's largest selection of Amplifiers, Recording equipment, etc., etc.



THE "TELETRON" BAND III CON-

Power pack kit for above if required— complete price 22/6 only. N.B. We are demonstrating this converter at Totten-ham Court Road.

demonstrating tins converter at return-ham Court Road. THE "SUPERIOR" FOUR KIT. Our new four-valve receiver. A.C. mains, 200/250 v. M. and Long waves. As with our very successful "Economy Four" all required components ar s X5GT and 6 vGCT. Cha 2 6 Seady dvilled. Cabinet size, 104 in. X 10in. wide, Maximum depth at base Sin. Lapering to 34in. at top. Sloping front. Very attractively finished in light wainut and peach. Each component brand new and tested prior to pack-ing. Complete instruction booklet with practical and theoretical dia-grams is provided. Booklet available at 1/6 post free. Our price for com-piete kit. \$69/96 I Please add 2/6 pack-ing adhent at 5/-, plus 2/6 packing and carriage. If preferred, we can supply Cabinet Assembly only, com-prising Cabinet and bracket wave-change switch, dial, pointer. drum pulley, drive spindle, drive spring and carriage. N.B.—Our kits are even sup-piled with sufficient solder for the job. unde specially wound Dence "Maxi-Q"



THE R.C. RAMBLER ALL-DRY PORTABLE KIT Full assembly details with practical and theoretical diagrams can be supplied at 1/6 post free. This is a truly professional 4-valve superhet —all dry—for medium and long waves. A cream plastic top panel, with dial engraved in red and green, with dial engraved in red and green. adds to the very imposing appear-ance of this model which is housed ance of this model which is housed in an attractive cream and grey leatherette covered attache-case type cabinet; measuring only 9in.x 7in.x 54in. Weight less batteries 42lb., with batteries 64lb. This set really has everything i Built-In frame aerial, high quality, extremely sensitive, and very adequate volume from the 5in speaker. Valve line-up specified, 185. IT4. Also the re-quired componens. exactly as specified, including cabinet, can be supplied from stock at the special inclusive price of £171/- plus 2/6 p. and p. (less batteries). Uses Ever-Ready 90 v. H.T. type B126 at 9/3. Also L.T. 1.5 v. A.D. 35 at 1/4.



RAMBLER MAINS UNIT! At last we are able to offer our special milns units kit for using our popular all-dry "Rambler" on A.C. Mains. Complete kit, wilch when assembled fits snug! nito battery compartment, can be supplied at 47/6, plus 1/6 packing and postage. Price includes all required components, and full assembly instructions. N.B.—This unit is completely self-contained in a metal box measuring fin. X24in.x 13In, and is ideally suitable for ANY, all-dry battery portable ANY all-dry battery portable requiring 90 v. H.T. and 1.5 v. L.T.



SUPER-OUALITY 6-VALVE RADIOGRAM CHASSIS Very limited quantity be Britaln's leading quality manufacturers, 3 wavehand, ECH42, L63, EF41, and EBC41. Com-bined pick-up amplifier and A.F. amplifier on Radio and Gram. Employs a special circuit for gramophone pre-amplification. Large glass dial horizontal tuning measuring lin. X3in. Chassis measure-ment: 144×9×81a. This is a superior chassis designed to sell originally in a Radiogram costing £79. Our price is 212/19/6 only, tax paid, plus 5/- packing and carriage. We will glady demonstrate this chassis or any other working item from our stocks, to personal callers I REGAL. A





VERY LIMITED QUANTITY.

MAINS TRANSFORMERS

Primary, 200-250 v. P. & P. 2/-300-0-300, 100 mA. 6 v. 3 amp., 5 v. 2 amp., 22/6.

Semi-shrouded, drop-through 380-0-380 v., 120 mA., 6.3 v. 4 amp., 5 v. 2.5 amp., 22/6.

Drop thro' 350-0-350 v. 70 mA. 6 v. 2.5 amp., 5 v. 2 amp., 14/6.

Chassis mounting or drop-thro'. Pri-110-250 v. Sec. 350-0-350, 250 mA. 6.3 v. 7 amp., 6.3 v. 0.5 amp., 5 v. O.T., 0.5 amp. 4 v. 4 amp. P. & P. 3/6 32/6

Chassis mounted and fully shrouded, 80 mA., 6 v. 3 amp., 5 v. 2 amp., 14/6.

250-0-250 80 mA., 6 v. 4 amp., 14/-.

Drop thro' 270-0-270, 80 mA. 6 v. 3 amp., 4 w. 1.5 amp., 13/6.

Drop thro' 270-0-270 60 mA., 6 v. 3 amp., 11 6.

250 v. 350 mA., 6.3 v. 4 a., twice 2 v 2 a., 19/6.

Auto-trans. Output 200/250 H.T. 500 v 250 m.A., 6 v. 4 a., twice, 2 v. 2 a., 19/6.

Auto-trans. Input 200/250. H.T. 350 v. 350 mA. Separate L.T.6.3 v. 7 a., 6.3 v. 14 anp., 5 v. 3 anp., 25/-P. & P. 3/-.

Pri. 200 v. Sec. 500-0-500 and 500-0-500 250 mA., both windings. 4 v. 3 amp. 4 v. 3 amp., 39/6. P. & P. 5/-.

Mains Transformer, fully impresented. Input 210, 220, 230, 240. Sec. 350-0-350 100 mA., with separate heater trans-former. Pri 210, 220, 230, 240. Sec. 6.3. v. 2 amp., 6.3. v. 3 amp., 4 v. 6 amp. and 5 v. 2 amp., 30/-. P. & P. 5/-

350-0-350 75 mA. 6.3 v. 3 a. tap 4 v 6.3 v. 1 a., 13/6.

500-0-500 125 mA. 4 v. C.T. 4 a C.T. 4 a., 4 v. C.T. 2.5 a., 27/G. 4 a., 4 v

500-0-500 250 mA. 4 v. C.T. 4 a., 4 v. C.T. 5 a., 4 v. C.T. 4 a., 39/6.

6}in. M.E. Speaker. 1,000 ohm field. 15/-.

R. & T.V. energised 641n. speaker with O.P. trans. field coil. 175 ohms 9/6. P. & P. 2/6.

R. & A. 61in. M.E. speaker, with O.P. trans. field 440 ohms, 10/6. P. & P. 2/6.

Volume Controls. Long spindles less switch, 50K, 500K, 1 meg., 2/6 each. P. & P. 3d. each.

Volume Controls. Long spindle and switch, $\frac{1}{2}$, $\frac{1}{2}$, 1 and 2 meg., $\frac{4}{2}$ each. 10X and 50X, $\frac{3}{2}$ 6 each. $\frac{1}{2}$ so at meg., long spindle double pole switch, minla-ture, $\frac{5}{2}$, P. & P. 3d. each.

Trimmers, 5-40 pf., 5d. 10-110, 10-250, 10-450 pf., 10d.

Twin-Gang .0005 Tuning Condenser, 5/-With trimmers. 7/8.

Twin Gang .0005, with feet, size $3\frac{1}{2} \times 3 \times 1\frac{1}{2}$ in., 6/6.

3-gang .0005. with feet, size $4\frac{1}{2} \times 3 \times 1\frac{1}{2}$ in., 7/6.

T.V. Coils, moulded former, iron-cored wound for re-winding purposes only. Ali-can 12 x 1 jin. J.V. each, 2 iron-core All-san 21 x jin. 1/6 each. The above coil formers are suitable for the "Wireless World" F.M. tuner.

Used Metal Rectifier, 250 v. 150 mA., 6/6. Metal Rectifier, 230 v. 45 mA., 6/-.

Metal Rectifier RM2, 125 v. 100 mA., 3/6

OUTPUT TRANSFORMERS. Standard OUTPUT TRANSPORMERS. Standard type 5,000 ohms imp.,4/8: 42-1 with extra feed-back windings, 4/8: Minia-ture 42-1, 3/3. Minit-rakio 3,500, 7,000 and 14,000, 5/6. 10-wat push-pul, 6% matching, 7/~, 90-1 3 ohm speech coll. 6/6.

STANDARD WAVE-CHANGE

SIADADA WAYE-URIVE SWITCHES. 4-pole 3-way, 1/9; 3-pole, 3-way, 3/6; 3-pole, 3-way, 1/9; 3-pole 3-way, 3/6; 3/6; Miniature type, long spindle 3-pole 4-way, 4-pole 3-way and 4-pole 2-way, 2/6 each. 2-pole 11-way, twin-water, 5/-; 1-pole 12-way single water, 5/-; F. 4, P. 34.



VALVE PUSH-PULL RADIOGRAM CHASSIS 7-VI A.U. MAINS 200/250v.
3 ware band, coverage short ware 16:50m, medium ware 187:550m, long ware 900-2,000m. 4 controls, volume control on off, tone control, tuning and wave change with gram position. Valve like up X79, W727, two DH77's, two EL 41's and EZ30. Output 7 watts. Size of chassis. 16in.X7in.X21n.Size of scale 21in.X41n.Overall height including hack plate 7% in. BRAND NEW. Fully guaranteed. P. & P. 7/6. £9/19/6.

PERMEASURE PERMEASURE VERTER for new commercial stallors. Input 300 ohm bai-anced line or 80 ohm coax. Coverage 180-200 Me/s. Vaive line-up 6AK5 RF amplifter.6AK5 mixer, d04 separate oscil-lator. This is a high Frequency coverage 60-

gain unit, ideal for fringe areas. Can also be used as EM TUNES. Frequency coverage 60-100 Mc/a. IF 10.7 Mo/s. Size 91n. wide, 61n. deep. 41n. high, 91n. scale, width, including over lap, 14in. Complete with S valves. P. & P. 3/-. £A/8/6. 10.7 Mc/s IF's to suit above, 4/6 each.

10.7 Mo/s IF's to suit above, 4/6 each. T.Y. CONVERTER for the new commercial stations, complete with 2 valves. Frequency can be set to any channel within the 186-196 M/s band 1.F will work into any existing T.V. receiver between 42-68 M/s. Input arranged for 80 ohm feeder, EFN0 as RF amplifier, EOCEI as local oscillator and mixer. The gain of the first stage, R.F. amplifier 10DB. Required power supply of 200 D.C. at 25 mA, 6.3, v.A.C. at 0.6 ampl. Input filter ensuing freedom from unvanted signals. Simple adjustments only, no Instruments required for trimming. Will work into any T.R.F. or superhet. Size 44 x 24 x 21. R. A. P. (26. £21)66. Bouble wound mains transformer, 200/250 v. metal rectifier, and smeothing condenser to suit above, 186. D. F.W.G. OCCUL 4.00 M. with FWA is suffisher mining dimension 15.1.

R.F.E.H.T. OSCILLATOR COLL. 6-9 K with EVS1 recisition reconcenser to suit above, 18/6. As above but complete with 6V6, EV51 and associated resistors and condensers. Circuit diagram, $15'_{-}$, $37'_{0}$. The above out completely built and tested in metal box. size 5 × 5 × 4§in., 42/6. I', & P. 3'_-

PLAP, M.-PLAP, M.-PLASTIC CABINET, as illustrated, 114 m. × ifln. × 54 m., in Winnut and Orean, also in polished Wainut complete with T.E.F. chassis, 2 wavehand scale, station names. new wave-band, hack-plate, drum, polnter, spring, drive spindle, S knobs and back, 22/6, P. & P. 3/6, AS ABOVE, with superhet chassis, 23/6, P. & P. 3/6, Either of the above items complete with 5h, P.M. speaker and O.P. transformer, 17/6 estra. Used metal rectifier, 230 v. 00 mA. 3/8; gang with trimmers, 6/8; M. and L.T.R.F. colfs, 5/-3 0 sholte ex-Gort, where, 3 v/h and chrcuit, 4/6; estate, the; own; whenge contch, 2/-1 52 x 33 mfd. 4/- bins conden-ser, 1/-; resistor kit, 2/-; condenser kit, 4/-Used A.C. mains 200/250 voits, 4 vaive plus n

SPECIAL OFFER.



Used A.C. mains 200/250 volts, 4 valve plus metal rectifier, medium wave superhet in polished waland cabinet, size 14 × 9; × 7µin., complete with valves 6K8, 6K7, 6Q7 and 6F6. 6; P.M. speaker. Fully guaranteed. P. & P. 7(6. 23/15/-.

P.M. SPEAKERS. 6jin. closed field, 18/6. 8in. closed field, 20/6. 10in. closed field, 25/-. 3jin., 16/6. P. & P. on each 2/-. EXTENSION SPEAKER in polished wainut, complete with Sin. P.M. P. & P. 3/-. 24/6.

B.S.R. MONARCH three speed automatic changer, current model, Brand new. Will take 7in., 10in, or 12in, records mixed. Turnover crystal head. Cream finish. A.C. Mains 200/250. £7/15/-. P. & P. 3/-.

1.2001t. High Impedance Reording Tape on aluminium spool. 12/6 post paid. OFB one-sixth h.p. A.C. 220/230 v. by Brook Motors. Reversible for continuous running. £2/18/6. Post and pkg. 7/6.

Radiogram Ghassis, 5 valve A.C./D.C. 3 wave-band superbet 195/255 v. 19-49, 200-556 and 1,000-2,000 metres, I.F. 470 K.c. size of chassis 13 x 6 $\frac{1}{2}$ x 2 $\frac{1}{2}$ I.n. size of scale 7 $\frac{1}{2}$ x 3 $\frac{1}{2}$ in. Valve inc-up 1801, 1809; 101:D11, U404 and 19P14. Twin mains filter input, 2 dial lights and Sin. P.M. 282/1706. F. & F. 5/-.

SPECIAL OFFER Sin. P.M. speakers, removed from chassis, fully guaranteed. All by famous manufacturers. P. & P. 1/6. 12/6.

CONSTRUCTOR'S PARCEL, medium and long wave A.C. mains 230/250 2-valve plus metal rectifier, comprising chassis $10\frac{1}{2} \times 4\frac{1}{2} \times 11$. 2 wave band scale, tuning condenser, wave-change swikely, volume control, heater trans., metal rectifier, 2 volves and viholders, smoothing and bias condensers, resisters and small condensers, and medium and long wave coll, litz wound, 22/6. P. & P. 2/6 etra. Circuit and point-to-point, 1/3.

wound, 22/6. P. & P. 2/6 ertra. Circuit and point-to-point, 1/3. CONSTRUCTOR'S PARCEL, comprising chassis 12 \pm x x 2 \pm in, cud, plated, 18 gauge, v/h., LF. and trans. cut-oute, back-plate, 2 supporting brackets, 3 wave-band scale. new wavelength stations names. Bize of scale 14 \pm x 4 \pm 11... dive, sp. drum. 2 publicys, pointer, 2 bulb holders, 5 pax. 1.0. v/b., 4 knobs and pair of 465 LF.s, twin gang, 16 x 16 mfd. 350 wkg., mains trans. 250-0250 60 mA. 6.3 v., 2 amp., 5 v. 2 amp. and 6 \pm in. M.E. speaker with O.P. trans., 39/6. P. & P. 3/6.

CR100 Coll packs in first-class condition less oscillator section, complete with 4-gang tuning condenser, 19/6. P, & P. 3/6.

CR100 465 Ke, I.F.s. types 3, 4 and 5 and F.B.O., new condition, 7/6 each. 465 Ke. Xtal lor CR100, 12/6

4-yang tuning condenser for CR100, 9/6.

POLISHING ATTACHMENT for electric drills. Quarter incb spindle, chromium plated 5in. brush, 3 polishing cloths and one sheepskin mop mounted on a 3in. rubber cap. Fort and pkg. 1/6. 12/6. Spare sheepskin mops, 2/6 each.

POTATO AND VEGETABLE FEELER. By Ismous manufacturer. To suit models A200 and A700, Capacity 41b. complete with water pump. All aluminium construction, white stove-enamelled finish. Originally intended for adaptation on an electric food-mixer, can be easily converted for hand operation. 39/6. P. & P. 3/-.

40-WATT FLUORESCENT KIT, A.C. mains 230/240. Comprising choke, power-factor con-denser, 2 tube holders, starter and starter-holder. P. & P. 3/-. 17/6.

20-WATT A.O. or D.C. 200/250 v. FLUORESCENT KIT comprising trough in white stove enamel thish, two tube holders, starter and holder and barretor. Post and packing 1/6. 12/6,

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Terms of Business : Cosh with order. Despatch of goods within 3 days from receipt of order. Where post and packing charge is not stated please add 1/6 up to 10/-, 2/- up to £1, and 2/6 up to £2. All enquiries S.A.E., lists 5d. each.

Mains Droppers. 0.3 ampt, 460 ohm tapped 280 and 410, 1/6; (2 amps, 4ohms, tapped at 100 ohm. 1/6; 0.3 amps, 950 ohms, tapped 740 and 825, 2/6; 0.2 amps, 1.000 ohms, vitreous, tapped 2/6; vitreous, 0.3 amp., 700 tapped 680, 640, 600, 31/6. F, & F, on each 34. TV Width Controls, 3/6

PERSONAL SHOPPERS ONLY. 91n. Enlarger. 17/6: 121n. 27/6. Germanium Crystal Diode, 1/6, post paid. Used 9in. Tube with ion burn. 17/6. Post paid.

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Speaker Matching Unit on aluminium chassis. 3-15 ohms reversible, 12/6. Line and E.H.T. Transformer, 14 Kv., using ferrocart core, complete with line and width control, and corona shields U37 rectifier winding, 35/-. The and 's and 's and control, and corns shields US7 rectiler winding, 35/-. Line and E.H.T. Transformer, 9 Kv., using ferrocart core, complete with built-inline and width control. Mounted on small all-chassis. Overall aize $4\dagger \times 1$ [in. EV51 rec. winding, 27/6. Scan coils, low line low impedance frame, complete with frame transformer to match above, 27/6. P. 4 P. 3/-. Line and E.H.T. Transformer, 9 Kv. ferrocart core, EV51, heater winding. complete with scan coils and frame output transformer, and line and width control, 35/-. P. 4 P. 3/-. As above, but complete with line and frame blocking transformers, 5 Henry 25 mA. choke, 100 mid. and 150 mid. P. 6 P. 3/-.

250 wkg. soo may be a solution of the solution 1/6 each. 32 mfd. 350 wkg.

32 mfd. 350 wkg. 16 × 24, 350 wkg. 4 mfd., 200 wkg. 16 × 8 mfd., 500 wkg. 16 × 16 mfd., 500 wkg. 25 mfd., 25 wkg. 250 mfd., 25 wkg. 250 mfd., 12 v. wkg. 16 mfd., 500 wkg., wire ends. 8 mfd., 500 v. wkg., wire ends. 8 mfd., 500 v. wkg., wire ends. 50 mfd., 25 v. wkg., wire ends. 10 mfd., 500 v. wkg., wire ends. 10 mfd., 500 v. wkg., wire ends. 10 mfd., 500 v. wkg., wire ends. 10 mfd., 250 v. wkg. 2/-1/3 5/9 3/9 4/-11d. 1/-3/3 2/6 1/6 1/9 100 mfd., 350 wkg. 100 mfd., 450 v. wkg., 280 mA., ā 100 mdd, 450 v. wkr., 280 mA., A.C. rippie 150 mdd, 350 v. wkr., 280 mA., A.C. rippie 200 mdd, 275 wkr. 16 + 16 mdd, 350 wkr. 50 mdd, 180 wkr. 65 mdd, 180 wkr. 8 mdd, 180 wkr. 60 + 100 mdd, 220 wkr. 50 mdd, 12 wkr. 50 mdd, 50 wkr. 3/11 4/8

17/6

Frame Oscillator Blocking Trans., 4/8. Line Osc. Blocking Trans., 4/6.

Line Oso. Biocking Trans., 4/6. CHOKES: 2-20 Men. 150 mA., 15/-. P. & P. 3/-. 6 Hen. 275 mA., 15/-. P. & P. 3/-. 100 Hen., 40 mA., 15/-. P. & P. 3/-. 2 heary 160 mA., 3/6; 250 mA. 10 henry, 10/6; 5 heary 200 mA. 60 ohms, 3/6.

being, 3(3), 10(6), 5 heiny 250 mA., 60 ohms, 3(3). Wide Anzie P.M. Focus Units. Vernier adj. state tube. 15/-. P.M. Focus Unit' for Mullard tubes with vernier adjustment. P. & P. 2/-. 15/-. Energised Focus Coll. low resistance mounting bracket, 17/6. Ion Traps for Mullard or English Elec-tric tubes, 5/-, post paid. Stardard 465 Kc. iron-cored I.F.s. 4 x 1] x 1] a., per pr. 7/6. Wearite standard, iron-cored, 465 Kc. I.F.s. 3 x 1] x 1] a., per pr. .9/6. Iron-Cored 465 Kc. Whitle Filter, 2/6.

(ron-tor(d 465 Kc, Whistle Pilter, 2/6. 465 KG, MIDGET LF.s. Q-120 size 1jin. iong, lin. wide, Hin. deep by very famous manufacturer. Pre-aligned adjustable iron-dust cores, per psir, 12/6.





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VOVEMBER, 1955

The TWIN 20 This is a complete fluorescent lighting. It has built-in ballast and ready to work. It is an ideal unit for the work-bench, and n shillar locations. There with two tubes, Post and insurance 2/6. b

THE TWIN 20

39/8.

RECORD PLAYER FOR £4/10/-.



3-SPEED INDUCTION MOTOR

3-speed motor with metal turntable and rubber mat. Latost rim drive with speed selection by knob at the shiel No auto, stop, but there is a stop position on the selector. Small mod. makes speed variable for special effects and dance work.

HI-FI PICK-UP

Using famous Cosmocord HI-O turnover crystal. Separate sapphire for each speed. Neat bakelite case with simple adjuster for weight compensation.

SPECIAL SNIP OFFER THIS MONTH

The two units for $\pounds 4/10/$ -, or 30/- deposit and four payments of 18/-, post and insurance, 5/-.



MINI-RADIO Uses high-efficiency coils-covers long and medium wave-bands and fits into

AMAZING



MULLARD AMPLIFIER " 510 "

A High Quality Amplifier designed by Mullard engineers. Robust high fidelity with a power output exceeding 10 watts and a harmonic distortion less than '4° at 10 watts. Its frequency response is extremely wide and level being almost flat from 10 to 20,000 C.P.S.--three controls are provided and the whole unit is very suitable for use with the Collaro Studio and most other good pick-nps. The price of the auth completely mail ou and ready to work is £10 the unit completely mail ou and ready to work is £10 the, plus 10/- carriage and insurance. Alternatively, if you wish to make up the unit yourself we shall be glud to supply the components separately. Sond for the Mullard amplifier shopping list.

A WONDERFUL CHRISTMAS PRESENT

Non-Mains, absolutely safe Children of all ages enloy playing Children of all ages enjoy playing records and will be overlayed to own the fine portable illustrated alongside. This uses the Garrard spring motor and a 2-valve battery amplifiar. The case is in two-tone initiation erveosdilationed skin. Special price £7/19/6, carriage 7/6 extra

THE ELPREQ ADDITA MIDLANDS ADDITA NOW AVAILABLE

ADDITA NOW AVAILABLE Our ADDITA Band III convertor which is enjoying such a hura sanossi ni the South is now available for Midlands viewers who will have heard that test transmissions are about to commence. Please be advised and order early, price is as for London model.

for London model. BETTER THAN FROM FACTORY BUILT SET Other constructors report better results from the convertor than from frckory-buil t Band III tilevisors. At Eastbourne, one of the latest models by a very famous maker would not receive the commercial signal on its own proper channel circuit despite trimming. However, with the ADDITA a reasonably clear and loud signal was received on Channel I without any adjustment.

ADDITA & T.R.F. TELEVISORS

Certain technicians have been of the opinion that our ADDITA would not be suitable for T.R.F. receivers. Full power transmissions seem to have disproved this theory, however, for we have beard from many viewers with the Viswmaster and other T.R.F. sets that they are getting good results For instance, in twing's post we heard from Mr. L. Campling, of Tolworth, Surrey, as follows:-"I would like to inform you that I have tested one of your convertors on both subject Band I and on Band III aerials. The reception was 100%. These test were acriates. The reception was 100%, These tests were carried out on inw power conditions and the acrisis were only about ten feet from ground level. The set was the home-made 'Electronic.''

AVAILABLE AS A KIT OR READY BUILT

AVAILABLE AS A KII OK KEAUY BUILT The price of the complete kit to build the ADDTA locinding valves, really wound coils, drilled and prepared chassis, handsome stove-enuanelled case, in fact every-thing, including transfers to decorate the front and identify the controls is 45/5/s- of grains com-ponents are also required. Post and insurance is 2/6 in each case. Data is included free with the parts or available separately price 2/6. When ordering please stabe whether for Midnawk or London area. Made up models for either area available price 27/10/-, plus 3/6 post and insurance post and insurance

"WIRELESS WORLD" BAND III CONVERTOR

One of the most successful circuits for Band III con-version was published in the "Wireless World," May 1934 We offer a complete kit of parts including the specified IF80 valves, wound coils, inflied chassis, in fact, everything including a copy of the cir uit diagram. Price ouly 42/6, post 2/6 extm. Maius components, is required, 25/- extra.

BAND III AERIAL KIT

An interesting serial, "The Polded V," was described in the July number of a T.V. magazine. We tried this and found it to be most efficient. It is simple to make. Kit comprises alloy elements and connectors, plastic centre piece and saidle for mounting. Price 8(8, plus 1/- post.

BAND III AERIALS

S-element array with swan-neck mast with "U" bolt clamp for fitting to exist-ing masts from in. to 2in. dia. AERIALS

21n. dia. 3-element array with cranked mast and wall mounting bracket..... 3-el-ment array with crauked mast and chimney tashing equipment 41/6

42/6

65/-

83/6

5-element array with swan-neck mast and "U" bolt clamp for fitting existing mast from §in. to 2in. dia. 52/6

> 53/6 671-

69/-

8-element array with cranked mast and chimney lashing equipment 8-element array with 15in. mast cap, 10ft. mast and heavy duty sincle chimney lashing equipment 10-element array with cranked mast and chimney equipment.....

10-element array with 15in. mast cap, 10ft. mast and heavy duty single chimney lashing equipment 145/-10-element array with 2in, mast cap, 12it. mast and heavy duty double chimney lashing equip-ment



This one Illustrated is the "E pares," it is undo ibtedly a beautiui piece of furnifure. It is elegantly veneered externally in figured section is mined to convenient level but is main event in the system of the system of the start cut. The lower deck acts as the motor huari, sgain is uncut, it measures 16×14 and has a clearance of 51n. from the lid. There is a compartment for the storage of recordings. overall dimensions of this essentially modern cabinet are

Sft. wide, 2ft. Slu, high, and lft. 411n. deep, Price £14/14/-, carriage and Insurance 20/-,



THIS IS ON OFFEL AT APPROX. HALF COST TO MAKE

An impressive costly look ing cabinet originally designed for T.V but simple modification makes the cabinet suitable for radiogram, ampliller, tape re-corder, or reflex speakersize 23in, wide, 22in, deep and 374in, high. Limited quantity at £8/15/- each, arriage 12/8.



CORNER CONSOLE

A massive cabinet but being corner fitted is not out of the even Is a proders mail living roots of this cabinet are 47in. Identified a state of this cabinet are 47in. Identified a state of the bouse 15in. Tele-visor, Rulio Unit, AmpHilter, Tape Deck, etc. Origin-ally 218. Our Frice 210 plus 30/-carriage. even In place



F.M. TUNER

This tuner is based upon the very successful circuit in the 134/booklef published by Data Publications. We have made up models at all branches and will be glad to demonstrate-94/6 Cost of all parts including valves, prepared metal chassis, wound colls and stove enamelies scale, slow motion drive, pointer, tuning knob, in fact everything needed to make the complete unit, is $\pounds 6/12/6$. Data is included free with the parts or is available separately price 2/-. 178/6



THE INFRAY MAJOR



This is a new type of over-head heater which in the main warms only the area within its radiant rays, and by so and by ao duing effects a very con-saving of fuel. One user in fact claims that in his office he re-ceives more heacht from the infacy Major than from a stan-from a stan-from a stan-thre times as much to run. duing

Perhaps one of the best points about The Infray Major is that its benefits are full numericharly, there is no warming-up period. It is essentially a personal type of heater, having its controls within easy reach of the operative. The control give four variations of heat and "off." At maximum heat the unit consumes 1 kW, which means that overall cost of heating can be controlled at 1d. per hour per operative (based vs the average cost of electricity, 1d. cer unit).

ld. per unit)

The Infray Major is of particular user-

- ne initial solid provides and solid provid

- contracts where spherically use functions to be a set of the set o

The unit has its uses in production for warming, drying, processing, etc. One special use being for warming through a slass screen, where for some reason or another the sub-stance cannot be warmed in free atmosphere. Price is £6/10/-, plus carriage and insurance 10/-.

COMMUNICATIONS RECEIVER R1155

This set is considered to be one of the fluest communications receivers available keday. The frequency range is 75 keys to 18 Mole. It is complete with 10 valves and is fitted in a black metal case. Used but completely overhauled and guaranteed in perfect working order. PRICES: Grade 2 \$7/19/8. Grade 1 £9/19/8. Or will be sent against deposit. If you cannot call to collect please include an additional 167- to cover cost of transit and case carriage.

MAINS POWER PACK FOR R1155

With Pentode output stage. Plugs into socket on receiver so no laternal modifications are required. Price $\xi_5/10/$ -complete with speaker ready to work, carriage 3/6. Or in cabinet $\xi_6/15/$. Carriage 5/-.

BENDIX RA-IB COMMUNICATIONS RECEIVER



Band 5 3.7 to 7.5 mc. The sensitivity is 4 micro-volts for full output. It uses 8 valves and operates from batteries (12 or 24 volt) of from the mains through a power pack. It has built-in output stage with a jack socket for phones. Controls, all or which are brought to the front panel, include: aerial switch, aerial compensating condenser, main tuning condenser, band selector. C.W. switch, power onoff switch, and volume control Very compactly built in enclue finished case, these sets are brand new having acever bean used al 10-or more - earing ord manacolid. Circuit dilate setarily and the setarily price 2/6, post free. given f

Mains Power Pack for Bendix RA-1B, £3/10/-

A NEW APPROACH

to an almost universal problem

An electronic computor to indicate football results [" the subject of our latest publication. The computor uses 3 valves and information is fed into 12 ratio arms. The result home, 'draw,'or 'away's indicated on a centre zero-meter, suitably scaled. The information to be fed into the ratio arms can be derived from the operator's own put mathods or without

The information to be fed into the ratio arms can be derived from the operator's own per methods, or alter-matively the data freely available in newspapers and popular magazines can be used. The circuit is guite an interesting one but is not at all difficult to construct or operate. When complete the unit is entirely self-contained and mains operated. The price of all components needed, excluding metal and chassis, is £310°, plus 3/6 post. Our Publication "A New Approach" is given free with orders, alter-natively will be supplied separately at 2/6, post free,

ENTIRELY NEW CIRCUIT

Redesigned and now built by the Cleveland Company-very good reports received.

THE "WINDSOR S"



This is a 5-valve A.C. superhet covering the usual long medium and short wavebands. It has a particularly fine clear dial with an extra long pointer travel. The latest type loctal raives are used and the chastle is complete and ready to operate. Chassis wire 159 fox din. Price £9/19/6, complete with Ain. speaker. Carriage and insurance 10/- E.P. terms if required.

FINSBURY TAPE RECORDER

This is a fine instrument using the now-famous Truvox Model TR7U tape deck, in conjunction with a 4-valve amplifier specially designed for tape work.

and play-back of the highest fidelity.

Its performance issuperiortomost proprietary re-corders of similar price level and as good as many marketed at much higher prices

It will take all standard tapes up to 1,200ft, providing up to two hours playing. It will also play new pre-recorded tapes.

The instruments are carefully checked before dispatch and can be heard working at any branch. Price, complete and ready to work, £43. plus £1/10/-carriage and insurance, partly returnable.

FOR HOME CONSTRUCTORS wishing to build equipment into their own cabinets the amplifier and tape dock are available separately, prices as follows:---

insurance.

P.V.C. HEATER WIRE

P.V.C. HEATER WIRE This has a resistance of 16 ohms per ft. It is wound on non-bygroscopic insulation and covered over with P.V.C abrunk sizeving. Quite mittable for use underground or under water ideal also for twisting around pipes to stop freesting or to preheat liquid. Price J-per yant.



SELECTION OF CHASSIS THE "ARMSTRONG" FC48

Among high class radio chassis, the name Armstrong is probably the most famous, and their new model FC48 certaing lives up to tradition. It is virtually a 10-vaire circuit, for among its sight valves two double triodes are employed. Special features of this chassis are (a) 8 with output in a push-pull circuit with ample negative feed-back to ensure the highest fidelity. (b) provision for using F.M., e.g., power irrength out to sockets and indicator on disi: (c) luclependent bass and treble controls with visual indica-tion of setting; (d) four wave batic overling [6-51, 60-120, 19-550, and 1,000-2,000 uctres. The size of this chassis is $12 \times 9 \times 9$ in. Price £23/18/-, plus 7/6 carriage and insure.



5-Valve 3-wave band superhet covering long, medium and

short wave. Ownam uniniature valves are employed and low loss iron cored coils account for an excellent signal-to-noise ratio, Full A.V.O. is applied to both frequency changer and L.F.

stages. The output stage utilises variable negative feedback. A gram, position is provided and reproduction of records is particularly good. Chassis size is 12 x 7 x 7in,-scale size is 10; x 4;in.

Chassis size is $12 \times 7 \times 7$ in.—scale size is $10_1 \times 44$ in. This receiver has been tested in particularly difficult areas and its stability and noise rejection have produced excep-tional results. Price £11/10' or £3 deposit—carriage, etc., 7/6. THE "TREMENDO"

THE "TREMENDO" This one is really superb. It has a 7-valve circuit with 6 wats output, fitted with independent bass and treble controls. It is really an efficient R.F. circuit coupled to a high-fidelity samplifer. The chassis size is the same as the Organized evolve, and it is built to the same eracting specifica-tion as the Organizer. Price 215/10/r, carriage and packing 7/6. H.P. terms if required. Ditto but with for the od earlai coll. 216/10/r.



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NOVEMBER, 1955

UNUSED RADIO & COMPONENT BARGAINS



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Each outfit is despatched in transit case at the amazingly low price of £3/-/- plus carriage 10/-. If despatched without Transit Case, £2/10/- plus carriage 8/6.

REMEMBER!! 9 Standard Valves, 2 Standard 460 Kc. IF's. 3 relays, a motor generator and a host of useful equipment and for breaking up we have a few with soiled cases but otherwise perfect and complete at £2 each, plus carriage 8/6.

We are offering as New, Complete TR.1196 Transceivers, as illustrated. Outfit comprises, 6 valve Superhet, 3 Valve Transmitter, Power Unit and Relay Unit. All complete on Chassis. Present range 4-6.5 mc/s. and output 2 watts. Can be easily converted to cover 1.5 mc/s.-7 mc/s. and power output up to 8 watts. It has a most versatile Receiver which can be easily adapted to cover any band of frequencies from medium broadcast to 30 mc/s. The Transmitter range can also be easily extended and by simply adding 200 pf. condenser to tank circuit will cover 1.5 mc/s. Circuit and conversion details included with each unit.





U.S.A. ELECTRO-VOICE MOVING COIL MICROPHONES The demand for these microphones has been exceptionally heavy. We still have a few left. Send your order now to avoid disappointment. Price $\pounds 2$ ea., have a few left. postage 1/6.

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For Cathode Rav Tubes having Heater/Cathode short circuit or for C.R. Tubes with falling emission. Type A. Low leakage windings. Ratio 1:1.25 giving a 25% boost on Secondary.
2 volt
10/5 each Ditto with mains primarles 12/8 each Type B. Maine input 220/240 voits. Multi Output 2.4,6,8,7,5,10 and 13 volts. Input has two taps which increase output volts by 25% and 50% respectively. This transformer is suitable for most Cathode Ray Tubes. With Tag Panel 21/- each.
Tubes. With Tag Panel 21/- each. Type C. Low capacity wound transformer for use with 2 volt Tubes with falling emission. Input 230/ 240 wolts. Output 2-2;2:2;3-2 volts at 2 amps. With tag Panel 17/9 each. All Isolation Transformers are individually boxed, labelled and clearly marked with relevant data.
RESISTORS. All values. 10 ohms to 10 meg., $\frac{1}{2}$ w., 4d.; $\frac{1}{2}$ w., 6d.: 1 w., 8d.; 2 w., 1/ HIGH STABILITY. $\frac{1}{2}$ w., 1%, 2/ Preferred values 100
ohms to 10 meg. 5 watt 3 25 ohms-10,000 ohms
15 000 ohne_50 000 ohme 5 m 1/0:10 m 9/9
WIRE-WOUND POTS. 3 WATT LAB. GOLVERN ETC. Fre-Set Min. T.V. Type Kunuried Slotted Knob All values 25 ohms to 30 K. 3/- en. 60 K. 4/- Ditto Carbon Track 50 K. UWW EXT. SPEAKER to 2 Mex. 3/- CONTROL 100 3/-
All values 25 ohms to 30 Values, 100 ohms to 50 K., K., 3/* a. 50 K., 4/. 5/6. Ditto Carbon Track 50 K. W/W EXT, SPEAKER to 2 Meg., 3/ CONTROL 100 3/ O/F TRANSFORMERS, Heavy Duty 50 mA., 4/6. Ditto.
O/F TRANSFORMERS. Heavy Duty 50 mA., 4/6. Ditto. tapped primary, 4/9. Multratic, push pull, 6/6. Tapped small pentode, 5/8. Ministure 334, etc., 3/9. L.F. OHO KES 15/10 H. 80/45 mA., 5/-: 25/20 H. 100/120 mA., 11/6; 20/15 H. 120/150 mA., 12/6.
MAINS TRANS. 350-0-350, 80 mA., 6.8 v, tapped 4 v. 4 a., 5 v. tapped 4 v. 2 a., dittn 250-0-250, 2U
MIDGET TRANS, 250-0-250 v. 50 mA. 6.3 v. 2 a., 12/6. HEATER TRANS, Tapped prim., 200/250 v. 6.3 v. 11 amp., 7/6; tapped sec. 2, 4, 6.3 v., 14 amp., 8/6.
CRYSTAL MIKE INSERT. Farnous make, precision engineered. Size only 18 × ³ / ₁₆ ln. Bargain price 6/6. No transformer required.
EXT. L.S. Switched Socket, on/off and parallel switching, complete with plug. 2/2
COPPER PLATED AERIAL RODS. 1 × 12in. push fitting. 2/6 doz., p. & p. 9d. AERIALTE EARTH RODS, 4/ ALADDIN FORMERS and cores. in., 8d.; in., 10d.
 ALADDA FORDERS and COPE. 10., 50.; 10., 100.; SLOW MOTION DRIVES, Epicycli ratio 61. 2/3. TYANA. Midget Soldering iron. 200/220 v. or 230/250 v., 14/11. TYANA TRIPLE THREE. Complete with detachable bench stand. 19/6. 200/220 v. or 230/250 v.
able bench stand. 19/6. 200/220 v. or 230/250 v. NEW SOLON MIDGET IRON. 25 w., 22/ IDEAL FOR RADIO CONSTRUCTORS. 200/220 v. or 230/250 v.
NEW SOLON MIDET IRON, 25 w., 22/-, IDEAL FOR RADIO CONSTRUCTORS, 200/20 v. or 230/250 v. MIKE TRANSF, Ratio 50:1, 3/9 ea., new and boxed. MAINS DROPPERS, 3 x 14in. Adj. Sliders, 3 amp. 750 ohma., 4/3, 2 amp., 1,000 ohma, 4/3.
LINE CORD3 amp., 60 ohms per foot, .2 amp., 100 ohms per foot, 2 way 6d, per foot, 3 way 7d, per foot. LOUDSPEAKERS P.M. 3 OHM. 6m. 16/6. 6jin. 17/6.
Goodmans Elliptical 7 \times 4 18/8. 8in. 19/8. 10in. 25/ 6jin. Goodmans with trans. 21/ Plessey 12in. P.M., 3 ohm, volce coll, 30/ 8in energised 2.6 K. or 2 K. ohm
neid, tapped output trans., 24.6; iees trans., 21 ORVSTAL DIODE. G.E.C., 3/6. Circuit Book 1/ H.R. PHONES. (Hi-grade Amer.), 15/6 pr.
ohma, 4/3, .2 amp. 1,000 ohma, 4/3. LINE CORD3 anp., 50 ohma per foot, .2 amp., 100 ohma per foot, 2 way 6d, per foot, 3-way 7d, per foot. LOUDSPEAKERS P.M. 3 OHM. 6n. 16/6. 6jin. 17/6. Goodmanas Elliptical 7 × 4 18/6. 8in. 18/6. 10in. 25/- 6jin. Goodmanas with trans. 21/ Plessey 12in. P.M., 3 ohm, voice coil, 30/ 8in. energised 2.5 K, or 2 K. ohm field, tapped output trans., 24/6: less trans., 21/ CRYSTAL DIODE. 6.E.C., 3/6. Circuit Book 1/ HEADPHONES , (Hi-grade Amer.), 15/6 pr. HEADPHONES , 4,000 ohms., 15/6 pr. HEADPHONES , 4,000 ohms., 15/6 pr. COPFER ENAMEL WHEE . t^{10} . 16 to 20 s.w.g., 2/-; 22 to 28 s.w.g., 2/6; 30 to 40 s.w.g., 3/6.
SWITCH CLEANER Fluid, squirt sport, 3/9 tin. TWIN GANG TUNING CONDENSERS0005 mtd. midget with triamers 8/68 375 pf. midget less trimmers 6/65; .0005 Btandard size with trimmers and feet 8/-; less trim- mers 8/-; dito. soiled, 2/6.
RECORDING TAPE
1,200ft. on standard fitting 7" Plastic reels. Brand new boxed 17/6
VALVE HOLDERS, Pax. Int. Oct. 4d., EF50, EA50 6d. B12A, CRT. 1/3. Eng. and Amer. 4, 5, 6, 7 and 9 pin, 1/ MOULDED Maxda and Int. Oct. 6d., B70, B8A, B8G, B9A 9d., B70 with can 1/6, VCR97 2/6. CERAMIC. EF50, B7G, B9A, Int. Oct. 1/-, B7G with can 1/9, B9A with can 2/6. SPEAKER FRET. Woven Plastle TYGAN. Wainut tone, 121n. wide, 2/- per foot. Expanded metal, gold or sliver, 94in. x 121n., 2/- per place.
SPEAKER FRET. Woven Plastic TYGAN. Wainut tone, 12in. wide, 2/- per foot. Expanded metal, gold or silver, 9§in. x 12in., 2/- per plece.
2/6 BRAND NEWSUB MINIATURE VALVES 2/6 Ex Deaf Ald Apparatus.
RF PENTODES L.F. PENTODES 0.625 v. FIL 1.25 v. FIL DF70 DL72
505AX} RAYTHEON {503AX 507AX

RADIO





-								
	BAND 3 T.V. CO	ONVERTOR KITS						
	Suitable all T.V. makes. T.R.F. or Superhet. Ready wound colls, B.V.A. valves, all components,							
- 1	Suitable all T.V. makes. T.R.F. or Superhet. Ready wound colls. B.V.A. valves, all components, punched chassis, circuit diagram, winnig plane, COMPLETE KIT for mains operation 200-250 v.							
	A.C. £3/10/ AS ABOVE less POWER PACK. Requires 200 v.							
į	20 ma H.T., 6.3 v. 1 a. L.	T., £2/5/						
	PUNCHED CHASSIS and WOUND COILS, com- ponent list, circuit diagram, wiring plans, only 19/6.							
	T/V PRE-AMP, Channel 1	. Midget Chassis 44 × 21 × valve, coaxial lead, Belling eady for use. Brand new.						
	plug and Octal plug Re Special Price, 21/	eady for use. Brand new.						
s .		1 20 ohm Coovial						
	Volume Controls	CABLE COAXIAI						
.C.	Long spindles. Guaran- teed 1 year. All values 10,000 ohms to 2 Meg. No Sw. S.P.Sw. D.P.Sw. 3/- 4/- 479	Seml-air spaced Poly- thene insulated. in. dla.						
am. sais	No Sw. S.P.Sw. D.P.Sw.	Losses cut 50% 9d.yd.						
tal bbs,	EXT. SPKR. TYPE 3/-	tin. Coaxial. 7d.yd.						
seis out	SOURCES 11							
1	T.V. AERIALS. Band I, a	Il channels in stock. Indoor						
IS	BAND III, indoor type, 7/6. BALANCED TWIN FEEDER	R per yd. 6d. 800 or 3000 ED FEEDER 9d. yd. 80 ohma. 9d. per yd. 4hn. dia. 0, 70 pf., 9d. 100 pf., 130 pf. 50 pf. 1/9. wrs. single or stranded 2d. yd. RS. 44d. each. 2EWDRIVER, 5/6.						
	50 CHM COAXIAL CABLE S	ED FEEDER 9d. yd. 80 ohma. 8d. per yd. 4in. dia.						
	1/3; 250 pf., 1/6: 600 pf., 72	50 pf., 1/9.						
- 1	5in. RADIO SCREWDRIVEI	RS. 41d. each.						
	VCR97 TESTED	FULL PICTURE £2						
61	MULTICORE SOLDER, 60/40	0, 18 s.w.g. 3d., 16 s.w.g. 4d.						
5	yard, I lb. 2/6. ALL BERNARDS BOOKS IN	STOCK.						
	Ditto, 12.5 kV., 9/6; 2 pf. t	to 500 pf. Micas, 6d.; Tubular						
6	.1/350 v. 9d.: .1/600 v. 1/3: CERAMIC CONDENSERS 50	ALL BERNARDS BOOKS IN STOCK. CONDENSERS. New stock001 mtd. 7 kV. T.C.C., 5/6. Ditto, 12.5 kV., 9/6; 2 pf. to 500 pf. Micns, 6d.; Tubular 500 v001 to .01 mtd., 9d.; .05 .1, 1/-; 23 1/6; .5 1/9; .1/350 v. 9d.: 1/600 v. 1/3:.1/1.500 v. 3/6. CERAMIC CONDENSERS, 500 v. 3 pf. to .01 mtd., 1/-, SILVER MICA CONDENSERS, 10%, 500 v. 5 pf. to 500 pf. 1/-, 600 pf. to 3,000 pf., 1/3.						
	SILVER MICA CONDENSER pf., 1/-, 600 pf, to 3,000 pf,	LS. 10%, 500 v. 5 pf. to 500						
Jnit lead	DITTO 1%, 500 pt. to 3,000 pt., 1/3, DITTO 1%, 500 v. 1.5 pf. to 500 pf., 1/9. 515 pf. to 5,000 pf., 2/							
rds. ‡in.								
ree.	A SMALL SELECTION FROM OUR STOCKS							
vith lays		3/6 3D6 <u>New & Guaranteed</u> 2/6 5/6						
top.	6J5 C10	H6M 2X2 6AG5 7/6 7/6 6B8						
rice	EB91 0Z4 6	6F6 12BE6 EC91 10/6						
ER.	6AC7 1T4 65	5K8 EF39 6AM6 SA7 EF92 6J6						
2in. 9/6,	6C9 384 6	V6G HVR2A EF55						
	68N7 3V4 6 7S7 6BE6 6	3X4 U22 EF80 500 <th></th>						
	ECH42 6BW6 12 12/6 Send Stamp EY51 List. All B.	for Full Valve 12K8						
2	EY51 List. All B. 1951 Low Tay	for Full Valve V.A. Types at 25Z4 x Prices.						
8								
0	TUBULAR CAN T 2/450 v. 2/3 16/500 v 8/450 v. 2/450 v. 2/3 500/12 v	LITYPES NEW STOCK YPES CAN TYPES 7. 4/- 8+16/450 v. 5/- 7. 3/- 32+32/350 v. 4/6 BASE 32+32/450 v. 6/6						
2.	2/450 v. 2/3 16/500 v 8/450 v. 2/3 500/12 v 16/500 v. 4/- SCREW	. 3/- 32+32/350 v. 4/6 BASE 32+32/450 v. 6/6						
	25/25 v. 1/9 TYPE 50/50' v. 2/- 8/500 v.	$\begin{array}{cccccccccccccccccccccccccccccccccccc$						
		. 41- 1 1,000/0 4. 400						
	100/25 v. 2/- 16/500 v	18 s.w.g. Plain, undrilled,						
	ALUMINIUM CHASSIS folded 4 sides and riveted Strong and soundly co	18 s.w.g. Plain, undrilled, d corners lattice fixing holes.						
lesk 78	ALUMINIUM CHASSIS folded 4 sides and riveted Strong and soundly co 7 x 4in., 4/6; 11 x 7in., 6 10/6; and 18 x 16 x 3in.	18 s.w.g. Plain, undrilled, d corners lattice fixing holes. Instructed with 21 In. sides. (9:13 × 9in., 8/6:14 × 11 in. . 16/8.						
78 ever on	ALUMINIUM CHASSIS folded 4 sides and riveted Strong and soundly co 7 x 4in., 4/6; 11 x 7in., 6 10/6; and 18 x 16 x 3in.	18 s.w.g. Plain, undrilled, d corners lattice fixing holes. Instructed with 21 In. sides. (9:13 × 9in., 8/6:14 × 11 in. . 16/8.						
78 ever	ALUMINIUM CHASSIS folded 4 sides and riveted Strong and soundly co 7 x 4in., 4/6; 11 x 7in., 6 10/6; and 18 x 16 x 3in.	18 s.w.g. Plain, undrilled, d corners lattice fixing holes. Instructed with 21 In. sides. (9:13 × 9in., 8/6:14 × 11 in. . 16/8.						
78 ever on inal	ALUMINIUM CHASSIS folded 4 sides and riveted Strong and soundly co 7 x 4in., 4/6; 11 x 7in., 6 10/6; and 18 x 16 x 3in.	18 s.w.g. Plain, undrilled, d corners lattice fixing holes. Instructed with 21 In. sides. (9:13 × 9in., 8/6:14 × 11 in. . 16/8.						
78 ever on inal	ALUMINIUM CHASSIS folded 4 sides and riveted Strong and soundly co 7 x 4in., 4/6; 11 x 7in., 6 10/6; and 18 x 16 x 3in.	18 s.w.g. Plain, undrilled, d corners lattice fixing holes. Instructed with 21 In. sides. (9:13 × 9in., 8/6:14 × 11 in. . 16/8.						
78 ever on inal	ALUMINIUM CHASSIS folded 4 sides and riveted strong and soundly co 7 × 4in. 4/6;11 × 7in6 10/6; and 18 × 16 × 3in. SENTERCEI. RECTIFIERS, VOITAGES. 8×/85 2 kV 4 3.6 kV8/6; K3/80 4 kV. MAINS TYPE. RMI.125 v6 RM3, 120 mA5/9; RM4 235 FULL WAVE BRIDGE SE or 12 v. 1 amp. 8/9; 2a. 11, CHARGER TRANSFORMES for charging ab 2, 6 or 12 v	18 s.w.g. Piah, undrilled, d corners lattice fixing holes. onstructed with 21n. sides. (9:13 × 91n., 8/6; 14 × 11n. 16/6. 						
78 ever on inal X D, vith Coil	ALUMINIUM CHASSIS folded 4 sides and riveted strong and soundly co 7 × 4in. 4/6;11 × 7in6 10/6; and 18 × 16 × 3in. SENTERCEI. RECTIFIERS, VOITAGES. 8×/85 2 kV 4 3.6 kV8/6; K3/80 4 kV. MAINS TYPE. RMI.125 v6 RM3, 120 mA5/9; RM4 235 FULL WAVE BRIDGE SE or 12 v. 1 amp. 8/9; 2a. 11, CHARGER TRANSFORMES for charging ab 2, 6 or 12 v	18 s.w.g. Piah, undrilled, d corners lattice fixing holes. onstructed with 21n. sides. (9:13 × 91n., 8/6; 14 × 11n. 16/6. 						
78 ever on inal X2D, vith Coll [c/s. cir-	ALUMINIUM CHASSIS folded 4 sides and riveted strong and soundly co 7 × 410. 4/6; 11 × 71n. 6 10/6; and 13 × 16 × 31h. SENTERCEL RECTIFIERS. VOLTAGES. K3/25 2 kV. 4 x, 6 kV. 8/6; K3/50 4 kV MAINS TYPE. BMI, 125 v. 6 RMi, 120 mA. 5/6; RM4 235 CHARGES. K3/25 2 kV. 4 VMAINS TYPE. MI, 125 v. 6 RMi, 120 mA. 5/6; RM4 235 CHARGES. TRANSFORME FULL WAVE BRIDGES E CHARGES TRANSFORME for charging at 2, 6 of 12 v ACD HYDROMETER. Packed in metal case 7 × 14 H F. MIDGET CHOKES. 14 WAVEPER ANGE SWITCHED	18 s.w.g. Pian, undrilled, d corners lattice faing holes, nastructed with 2 10. sides. //9;13 × 91n. s/6;14 × 1110. 10/6. 						
78 ever on inal X 2D. vith Coil Ic/s. cir- nner	ALUMINIUM CHASSIS folded 4 sides and riveted strong and soundly co 7 × 410. 4/6; 11 × 71n. 6 10/6; and 13 × 16 × 31h. SENTERCEL RECTIFIERS. VOLTAGES. K3/25 2 kV. 4 x, 6 kV. 8/6; K3/50 4 kV MAINS TYPE. BMI, 125 v. 6 RMi, 120 mA. 5/6; RM4 235 CHARGES. K3/25 2 kV. 4 VMAINS TYPE. MI, 125 v. 6 RMi, 120 mA. 5/6; RM4 235 CHARGES. TRANSFORME FULL WAVE BRIDGES E CHARGES TRANSFORME for charging at 2, 6 of 12 v ACD HYDROMETER. Packed in metal case 7 × 14 H F. MIDGET CHOKES. 14 WAVEPER ANGE SWITCHED	18 s.w.g. Pian, undrilled, d corners lattice faing holes, nastructed with 2 10. sides. //9;13 × 91n. s/6;14 × 1110. 10/6. 						
78 ever on inal X 2D, vith	ALUMINIUM CHASSIS folded 4 sides and riveted strong and soundly co 7 × 4hc. 4/6;11 × 7hc., 6 10/6; and 13 × 16 × 3hr. SENTERCEI RECTIFIERS. VOLTAGES, K3/25 2 kV., 4 3.6 kV., 7/8; K3/50 4 kV MAINSTYPE, RM1, 125 v., 6 RM4, 120 mA., 5/9; RM4 256 FULL WAYE BRIDGE SE or 12 v. 14 amp. 8/9; 2a. 11 CHARGER TRANSFORMER for charging at 2, 6 or 12 v ACID HYDROMETER. Packed in metal case 7 × 14 WAVECHANGE SWITCHES. 5 p. 4-way 2 wafer. 2 p. 2-way, 4 p. 2-way, 4 p. 3 p. 4-way, 1 p. 12-way	18 s.w.g. Pian, undrilled, d corners lattice faing holes, nastructed with 2 10. sides. //9;13 × 91n. s/6;14 × 1110. 10/6. 						

307 WHITEHORSE ROAD, WEST CROYDON OPEN ALL DAY-(Wed. 1 p.m. 10 page list S.A.E. Phone THO 1665, after 6 p.m. 4198, Buses 133 or 68 pass door. 48-hour postal Service. P. & P., 6d. El orders post free. C.O.D. Service I/-

KNOB3, GOLD ENGRAVED. Wainut or Ivory, 1jin. diam., 1/6 each. "Pocus," "Contrast," "Bril-lance," "Brilliance — On-Oft," "On-Oft," Volume," "Vol.-On-Oft," "Tone," "Tuning," "Treble," "Bass," "Wavechange," "Radio Gran," "S.M.L. Gram," "Record-Play," Brightness," Ditto, not engraved, 1/- each.

NOVEMBER, 1955





Special Offer Brand New T-350XM TRANSMITTERS

These magnificent 350-watt Transmitters manufactured by the Technical Radio Co. of California, U.S.A., are offered at a fraction of their original cost.

Frequency Range: 2,000-20,000 Kc/s. Power Output: Radiotelegraph 350 watts. Radiotelephone 250 watts.

Frequency Control by Built-in Master Oscillator or Crystal Multiplier.

A.C. Mains 210/250 volts, 50-60 cycles.

Size: $58\frac{1}{2}$ in. high. Gross Weight : 16in. wide. 967 lbs. $24\frac{1}{2}$ in. deep. Shipping Capacity: 44.6 cu. feet

Comprising:-As Shipped In Manufacturers' Original Sealed Transit Cases.

Case I. Tr	ansmitter	Case 2.	Speech	Amplifier
72	5 lbs.		82 lbs.	
36	cu. feet		4.7 cu.	feet.
Case 3.	Low Voltage	160 lbs.		
	Power Deck	3.9 cu.	feet,	
	and Modulati	ion Tran	nsforme	r

The whole Transmitter complete with all Valves, full instruction books containing complete circuits, wiring and technical data, ready for immediate shipment. PRICE UPON APPLICATION.


NOVEMBER, 1955

PROOPS BROS. LTD. The Walk-around Shop 2-METRES!

RECEIVER TYPE R1392

FREQUENCY 95-150 Mc/s (2-3 METRES) **AIR TESTED 15 VALVE SUPERHET**

Valve line up: 1st and 2nd R.F. Amp VR.136 (EF.54); 1st local oscillator VR.65 (SP.61); 2 Oscillator Multipliers, VR.136 (EF.54); 3 I.F. Amp V.R.53 (EF.39); A.G.C. 607; Output 6J5; Muting VR.92; (EA50); Noise Limiter VR.92 (EA.50); B.F.O. 6J7; Mixer VR.136 (EF.54); De Mod. 6Q7.

Slow motion Tuning over 95-150 Mc/s or can be Crystal controlled.

Power supply required: 240-250 volts at 80 mA 6.3 volts at 4 amps. Size 19" × 10" ×10" Standard Rack Mounting. PRICE £6. 19.6. Complete with valves and circuit diagram, Packing and postage 17/6. 10/- returnable on packing case.

MAINS POWER UNIT TYPE 234 USED WITH R1392 and R1192

With change-over link. Double smoothed. Input 200-250 volts 50 cycles. Output 240 volts at 200 mA. 6.3 volts at 6 amps. with volt-meter 0-300v Reading Input and Output voltages. Size $19'' \times 10'' \times 6\frac{1}{2}''$ Standard Rack Mounting.



PRICE £3. 10. 0. Packing Postage 7/6 (limited quantities).



RECEIVER TYPE RII32

FREQUENCY 100-126 Mc/s. 11 VALVE SUPERHET Valve Line Up: R.F. Amplifier VR.65 (SP.61); Frequency changer VR.65 (SP.61): Local Oscillator VR.66 (P.61); Stabilizer VS.70 (7455); $3 \times I.F.$ Amplifiers VR.53 (EF.39); B.F.O. VR.53 (EF.39); Detector VR.54 (EB.34); A.F. Amplifier VR.57 (EK.32); Output VR.67 (6J5).

Switchable, A.G.C. and A.V.C. Variable B.F.O.

Circuit diagrams with units. Easily converted to cover Wrotham Band. No alterations to wiring required. Conversion Slugs and instructions. 5/- Extra. Size $19'' \times 10'' \times 10''$. Standard Rack Mounting. PRICE £3.7.6. Packing and carriage 15/-. 10/- returnable on Packing Case.

2/6 post paid

FL8 RADIO FILTERS with PL 55 and JK26, plug and socket. Size $2\frac{1}{2}$ in. $\times 3$ in. $\times 2\frac{1}{2}$ in. 10/6 each. 1/6 Post and packing **REFLECTOR** in bakelite case fitted with small bayonet cap holder. Size 5in. in diameter by 3in. deep.



Make a miniature POCKET RADIO

All these fine offers are on display at **m**



Incorporating high "Q" technique using the New Ferrite rod. Made possible by simple conversion of an ex-Govt. Hearing Aid.

Technical Details. A Germanium Diode Detector circuit followed by Technical Details. the existing 3 valve Amplifier, giving adequate amplification throughout the medium wave band.

This conversion can be carried out in approximately 30 minutes.

SEE and HEAR this Miniature **POCKET RADIO** demonstrated.

THE COMPLETE KIT OF PARTS includes a Type OL10 Hearing Aid (with Crystal microphone) in perfect working order with miniature ear phone and moulded ear insert attached; ferrite rod, germanium diode, components, circuit diagram and full instructions. Price £2 6s. Od. post paid

ALL COMPONENTS SOLD SEPARATELY Deaf Aid Unit with earpiece £1. 15. 0. Plastic Ear Mould Ferrite Rod 2. 0. 5. 0. Ferrite Rod Conversion Components Batteries 1.5v. L.T. (Type D.18) 30v. H.T. (Type B.119) 4. 0. 4. 3.

NOTE: As the crystal microphone is not used in the Pocket Radio, it can, if desired, be used as a general microphone and it does not require a matching transformer.





128

commemorate their Ist Anniversary with these offers



F.M.

10db.

ABSORPTION WAVEMETER

R.F. 24 20/30 Mc/s Switched

R.F. 25 40-50 Mc/s Switched

R.F. 26 50-65 Mc/s Variable

Packing and postage 2/- ea.

Tuning Valved (Dials Damaged)

Tuning Valved 9/6 ea.

Tuning Valved 9/6 ea.

15/-

R.F. UNITS

ELECTROSTATIC LOUDSPEAKERS

High fidelity Electrostatic Tweeters 3,000 to 20,000 cycles. This loudspeaker is essential when receiving

recordings and TV sound. 20,000 cycles response+

Type LSH 75 size $3'' \times 3'' \times \frac{1}{4}''$ Price **12/6** post paid.

Type LSH 100 size $5^{"} \times 4^{"} \times \frac{3}{4}^{"}$ Price 21/- post paid.

transmissions or reproducing high quality

Easily converted to 2 metres or 70 cm. In Copper-plated metal case $3\frac{1}{2}$ × $4\frac{1}{2}$ × $5\frac{1}{2}$ with dial calibrated 0-100 and 80V Neon Tube. Coverage approx. 190-210 Mc/s. New. 6/6 each post paid.

TYPE 62 INDICATORS Ideal for conversion to oscilloscopes, T.V. units, etc. Containing V.C.R.97, 12 VR.91 (EF.50), 2 VR.54 (EB.34), 3 VR.92 (EA.50), 4 CV.118. Slow-motion dial, 13 Pots and scores of useful components. Size: $8\frac{1}{2}'' \times 11\frac{1}{2}'' \times 18''$. New in wooden packing case. £3. 10. 0. carriage 7/6.



RECEIVER UNIT EX-TR1143A

Suitable for conversion to 2 metres and F.M. Wrotham. Circuit diagram free.

Price-less valves, D/- post paid.



AMPLIFIER UNIT Ex-TR1143A.--- A 3-stage transformer coupled amplifier. Push pull VT52s output to modulate push pull VT501s. Circuit diagram free.

TRANSMITTER UNIT Ex-TR1143A.—Suitable for conversion to 2 metres. Circuit diagram and coil conversion details supplied free. Price, less valves, 5/- post paid.

Price, less valves, 4/6 post paid.

A SELECTION OF EX-GOVERNMENT SURPLUS VALVES

1A3 4/6 ea. 6F7 6/6 ea. 6SN7 6/6 ea. FW4/800 9/- ea. VT52 (EL32) 5/- ea.	-
The LOC FIG HOAD BIG DENILS BI MERCAL (PETAL) FI	– ea.
1A5 4/6 ,, 6G6 5/6 ,, 6SQ7 7/6 ,, PEN46 7/- ,, VT501 (TT11) 5/- ,	- 37
3B24 2/6 " 6H6 2/6 " 6V6G 6/6 " VR53 (EF39) 5/- " VU111 2/- "	- ,,
5U4G 7/6 " 6H6 Metal 3/- " 6V6 Metal 8/6 " VR55 (EBC33) 7/- " 446A 25/- "	- ,,
6AC7 5/- " 6K6.GTG 5/6 " 6X5 7/- " VR56 (EF36) 5/- " 713A 8/6 "	6 ,,
6AUS 0/0 , 0L0 1/0 , 1215 5/- VR65 (SP61) 3/6 807 USA 6/6	6 ,,
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	- ,,
6C5 5/- " 6SJ7 7/- " 12SH7 Metal 5/- " VR91 (EF50) 5/- " 954 2/6 "	6 ,,
6F6.GTG 7/- ", 6SK7 5/6", 12U5 5/- ", VR135 5/- ", 9004 3/6",	6 "
All above prices post paid.	

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LONDON

BROS. LTD.

Plug and socket, 3 pin, waterproof, 5 amp. Suitable for caravan or trailer. 1/6 per pair post paid.

Jones type plug and socket 6-way. 1/9 per pair post paid. Selenium Rectifiers 125V, 80mA Chassis cooled 1 in. ×1 in. $\times \frac{1}{4}$ in. $\frac{4}{3}$ post paid.

Microphones E.M. with switch. Boxed, new 1/6 p.p. Alarm Bells, U.S. manufacture. 24V d.c., 3‡ in. dia. Stout aluminium casting. New, boxed 6/- p.p.

Rectifier Unit, containing half wave $12V \frac{1}{2}A$ selenium rectifier 1 in. dia. housed in Bakelite case, size $2\frac{1}{4}$ in. $\times 1\frac{1}{4}$ in. \times 13 in. 1/3 p.p.

Yaxley type Switch, Base mounting 2 pole, 5-way, 4 bank, 6 in. spindle. 3/- p. p.

Aerial Rods, 12 in. $\times \frac{1}{4}$ in. sections, copper plated. 2/3 per doz.

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52 TOTTENHAM COURT ROAD

Cathode Ray Tube type 3BP1, with mu-metal screen and base. 3 in, short persistence tube. 22/6 p.p. Hand Generator (ex-Dinghy Transmitter) 28V, 175A and

300V 40mA output. Containing useful reduction gearing, housed in strong aluminium casting. Can be used for hand

bench grinder, basis for megger, etc. Generator can be converted to mains motor. 15/- p.p. Mains Motor, 220-250V a.c./d.c. Spindle $\frac{3}{8}$ in. extending 1 in., 13/16 pulley with v groove at $\frac{5}{8} \times \frac{5}{32}$. Overall length $8\frac{1}{2}$ in., dia. $3\frac{1}{4}$ in. Base plate 4 in. square, 4 hole fixing. 15/- p.p.

3 Bank Toggle Switches, 5 amp. in Bakelite housing. Ref. 5C/544. 2/- p. p.

25-Way, 2 Bank Switch. Contained in metal case 9 in. ×6 in. $\times 2\frac{1}{2}$ in., with two 12-way terminal blocks. Useful for multi-test meters, model control, etc. Type 5D/460, 7/6 p.p.

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NOVEMBER, 1955



"MUST HAVE" BARGAINS!

BC.610 MODULATION TRANS FORMERS, Brand new spares for this famous American Transmitter. In Maker's Original Cases, ONLY £5 each.

RE UNITS TYPE 26. For use with the R.1355 or any receiver with a 6.3 v supply. This is the variable tuning unit which uses 2 valves EF54 and 1 of EC52. Covers 65-50 Mc/s (5-6 metres). Com-plete with valves, and BRAND NEW IN MAKER'S CARTONS. ONLY 29/6 each

"PYE" 45 MC/S I.F. STRIP. Ready made for London Vision Channel, this 5-stage strip contains 6 valves EF50 and I EA50. Supplied with circuit and details of very slight mods. required. BRAND NEW, ONLY 69/6 or less valves 50/-.

TRANSFORMERS. Manufactured to Upright mounting, fully shrouded,

Upright mounting, fully shrouded, normal primaries. 425 v.-0-425 v. 250 mA., 6.3 v. 4 a., 6.3 v. 4 a., 5 v. 3 a., 65/-. 350 v.-0-350 v. 160 mA., 6.3 v. 6 a., 6.3 v. 3 a., 5 v. 3 a., 47/6. 350 v.-0-250 v. 150 mA., 6.3 v. 5 a., 0-4.5 v. 3 a., 37/6. 250 v.-0-250 v. 60 mA., 6.3 v. 3 a., 5 v. 2 a., 21/-. Please add 2/- per transformer postage.

TRANSFORMERS, FILAMENT. 6.3 v. 2 a., 7/6, 6.3 v, 3 a., 10/6 (postage 1/-).

TRANSFORMERS, EHT. Upright

mounting. EHT for VCR97 Tube 2,500 v. 5 mA. EMT for VCR97 Jube 2,500 v. 5 mA. 2 v.-0-2 v. 1, 1 a., 2 v.-0-2 v. 2 a., 42/6. EMT 5,500 v. 5 mA., 2 v. 1 a., 79/6. EMT 7,000 v. 5 mA., 2 v. 1 a., 89/6. Please add 2/- per transformer postage.

TRANSFORMER EHT. Unrepeat-able "snip "250 v. primary, secondary 2,000 v. R.M.S. (approx. 2,800 v. D.C.). Size 2²/₂ x 2¹/₂ x 3¹/₂ in. for tag panel. ONLY 15/- (post, etc., 2/-).

TRANSFORMERS, EX W.D. AND ADMIRALTY, built to more than 50 per cent, safety factor with normal A.C. mains primaries. Brand new and unused, 330-0-330 v: 100 mA. 4 v. 3 a., 22/6.

L.T. HEAVY DUTY. Has 3 separate windings of 5 v.-0-5 at 5 amps., and by using combinations will give various voltages at high current. ONLY 39/6 voltages at high current. ONLY 39/6 Please add 2/6 per transformer postage.

SPECIAL OFFER. Ex-Admiraley L.T. TRANSFORMER. Normal mains input, output 4 v. 20 amps. C.T. New and unused, these have become damaged, but are still usable, the damage being confined to broken the damage being confined to broken fixing lugs, and/or broken bakelite terminal panels. Formerly sold at 30/-, now offered at 17/6 (post, etc., 2/6).

COMMUNICATIONS RECEIVER R.1155

The famous ex-Bomber Command Receiver known the world over to be supreme in its class. Covers 5 wave ranges: 18.5-7.5 Mefs., 7.5-3.0 Mefs., 1,500-600 kcfs., 500-200 kcfs., 200-75 kcfs., and is easily and simply adapted for normal mains use. Full details being supplied. Aerial tested before despatch. BOAND NEW AND UNUSED IN MAKER'S TRANSIT

75 kc/s, and is easily and simply adapted for normal mains use. Full details being supplied. Aerial tested before despatch. BRAND NEW AND UNUSED IN MAKER'S TRANSIT CASES. ONLY 611 19 6. BRAND NEW BUT SHOP SOILED, also tested working before despatch. 69:19 6 (carriage 10/6). A.C. MAINS POWER PACK OUTPUT STAGE, in black metal case, enabling the receiver to be operated immediately, by just plugging in, without any modification. Can be supplied as follows: WITH built-in 6§in. P.M. Speaker, £5/5/-, LESS speaker, £4 10/- (carriage 3.6). DEDUCT 10/- IF PURCHASING RECEIVER AND POWER PACK TOGETHER. Send S.A.E. for illustrated leaflet, or 1/3 for 14-page bookled which gives technical information, circuits, etc., and is supplied free with each receiver.

BAND III CRYSTAL CALIBRATORS Manufactured by Marconi Instruments. Frequency range 170-240 Mc/s. Has directly calibrated dial and incorporates 5 Mc/s. 240 Mc/s. Has directly calibrated dial and incorporates 5 Mc/s. crystalfor accuracy of better than .001%. Internal power pack for normal A.C. mains operation. Completely self-contained in grey metal case size 154in. x 10in. x 10in., and ready to operate, with spare set of valves, and instruction manual. In original maker's transit case, BRAND NEW. ONLY £5/19/6 (carriage, etc., 10/6).

RUNNING HOUR METERS

For checking running time of equipment up to 9,999 hours. Operates from normal 50 cycles mains. BRAND NEW IN MAKER'S CARTONS. ONLY 39/6.

"ALL DRY" RECEIVER MAINS UNIT

If your battery portable uses midget 1.4 v. valves of the 154-114-1R5-354 series, then this will save you pounds in bat-teries. Delivers LT and HT from normal mains. Manufactured by "AVO" for the Ministry of Supply. Fused on mains side, and ready to work. Size B#in.x5#in.x3fin. BRAND NEW. ONLY 39% (carriage, etc., 26). Covers 6.0-9.0 Mc/s.

F.S.D. SIZE AND TYPE PRICE Im.a. D.C. 34in. Flush circular 22.6 1 m.a. D.C. 34in. Flush circular (scaled 600 v.) 52.6 10 m.a. D.C.		METERS	
1 m.a. D.C. 3 jin. Flush circular (scaled 600 v.) 52/6 10 m.a. D.C. 2 jin. Flush circular (blank scale) 10 '6 150 m.a. D.C. 2 jin. Flush circular (blank scale) 10 '6 500 m.a. thermo 2 jin. Flush square 7/6 1 amp. thermo. 2 jin. Proj. circular 6/6 3 amp. thermo. 2 jin. Flush square 5/- 4 amp. D.C. 2 jin. Flush square 12/6 20 amp. D.C. 2 jin. Proj. circular 12/6 4 amp. D.C. 2 jin. Proj. circular 12/6 500 amp. D.C. 2 jin. Proj. circular 12/6 4 amp. D.C. 2 jin. Proj. circular 12/6 4 amp. D.C. 2 jin. Proj. circular 12/6 500-30 amp. D.C. 2 jin. Flush square 12/6 15 volts A.C. 2 jin. Flush circular moving iron 5/- 20 volts A.C. 3 jin. Flush circular moving iron 25/- 300 volts A.C. 2 jin. Flush square 25/- 300 volts D.C. 2 jin. Flush square 10/6	F.S.D. SIZE AND	TYPE	PRICE
600 v.) 52/6 10 m.a. D.C. 24 in. Flush circular (blank scale) 10'6 150 m.a. D.C. 2in. Flush square 7/6 500 m.a. D.C. 2in. Proj. circular 6/6 3 amp. thermo 2in. Proj. circular 6/6 4 amp. D.C. 2in. Flush square 7/6 20 amp. D.C. 2in. Proj. circular 6/6 20 amp. D.C. 2in. Proj. circular 7/6 20 amp. D.C. 2in. Proj. circular 7/6 30-0-30 amp. D.C. 2in. Flush square 12/6 50-0-50 amp. D.C. 2in. Flush square 12/6 15 volts A.C. 24 in. Flush circular moving 10'6 300 volts A.C. 24 in. Flush circular moving 25/- 300 volts D.C. 2in. Flush circular moving 10'6 300 volts D.C. 2in. Flush square 25/- 300 volts D.C. 2in. Flush square 25/- 300 volts D.C. 2in. Flush square 25/- 300 volts D.C. 2in. Flush square 10'6	I m.a. D.C.	21in. Flush circular	22,6
10 m.a. D.C.	1 m.a. D.C.		
10 m.a. D.C.		600 v.)	52/6
150 m.a. D.C. 2in. Flush square 7/6 500 m.a. thermo 2in. Proj. circular 5'- 1 amp. thermo. 2in. Proj. circular 6'/ 3 amp. thermo. 2in. Flush square 5'- 4 amp. D.C. 2in. Flush scrular 12'/ 20 amp. D.C. 2in. Proj. circular 12'/ 30-0-30 amp. D.C. 2in. Flush square 12'/ 50-0-50 amp. D.C. 2in. Flush square 12'/ 15 volts A.C. 2'//in. Flush circular moving 10'/ 20 volts A.C. 2'//in. Flush circular moving 2'/ 300 volts A.C. 2'// 2'/ 2'/ 300 volts D.C. 2in. Flush square 10'/ 2'/	10 m.a. D.C	21in. Flush circular (bl	ank
500 m.a. thermo 2in. Proj. circular 5 1 amp. thermo. 2iin. Proj. circular 6/6 3 amp. thermo. 2iin. Flush square 5 4 amp. D.C. 2jin. Flush square 5 20 amp. D.C. 2in. Proj. circular 12/6 20 amp. A.C. 2jin. Flush circular 12/6 20 amp. D.C. 2in. Proj. circular 12/6 30 amp. D.C. 2in. Proj. circular 12/6 30-0.30 amp. D.C. Car type moving iron 5 50-0.50 amp. D.C. 2in. Flush square 12/6 15 volts A.C. 2jin. Flush square moving iron 8/6 20 volts A.C. 2jin. Flush circular moving iron 25/- 300 volts A.C. 2jin. Flush square moving iron 25/- 300 volts D.C. 2in. Flush square moving iron 25/-		scale)	10.6
1 amp. thermo. 24 in. Proj. circular 6/6 3 amp. thermo. 2in. Flush square 5/- 4 amp. D.C. 23 in. Flush square 12/6 20 amp. D.C. 21 in. Proj. circular 12/6 20 amp. D.C. 21 in. Proj. circular 12/6 20 amp. D.C. 21 in. Proj. circular 12/6 40 amp. D.C. 21 in. Flush circular 12/6 30-0-30 amp. D.C. 21 in. Proj. circular moving iron 5/- 50-0-50 amp. D.C. 21 in. Flush square 12/6 15 volts A.C. 24 in. Flush circular moving iron 8/6 20 volts A.C. 34 in. Flush circular moving iron 2/- 300 volts A.C. 24 in. Flush circular moving iron 2/- 300 volts D.C. 21 in. Flush circular moving iron 2/- 300 volts D.C. 21 in. Flush circular moving iron 2/- 300 volts D.C. 2/- 2/- 10/6	150 m.a. D.C	2in. Flush square	7/6
3 amp. thermo. 2in. Flush square	500 m.a. thermo	2in. Proj. circular	5/-
4 amp. D.C. 2 jin. Flush circular 12/6 20 amp. D.C. 2 jin. Proj. circular 12/6 20 amp. A.C. 2 jin. Flush circular 12/6 40 amp. D.C. 2 jin. Flush circular 12/6 50-0.30 amp. D.C. Car type moving iron 5/5 50-0.50 amp. D.C. 2 jin. Flush square 12/6 15 volts A.C. 2 jin. Flush square 12/6 20 volts A.C. 3 jin. Flush circular moving 8/6 300 volts A.C. 2 jin. Flush circular moving 25/- 300 volts D.C. 2 jin. Flush square 10/6	I amp. thermo.		6/6
20 amp. D.C. 2ín. Proj. circular	3 amp. thermo.	2in. Flush square	5/-
20 amp. A.C. 24in. Flush circular 12'6 40 amp. D.C. 2in. Proj. circular 7'6 30-0-30 amp. D.C. Car type moving iron 5'6 50-0-50 amp. D.C. 2in. Flush square 12'6 15 volts A.C. 24 in. Flush circular moving 12'6 20 volts A.C. 34 in. Flush circular moving 16'6 300 volts A.C. 24 in. Flush circular moving 16'6 300 volts A.C. 24 in. Flush circular moving 25'- 300 volts D.C. 20 in. Flush square 25'- 300 volts D.C. 21 in. Flush square 10'6	4 amp. D.C.		12/6
40 amp. D.C. 2/in. Proj. circular	20 amp. D.C.		
30-0-30 amp. D.C. Car type moving iron 5/- 50-0-50 amp. D.C. 2in. Flush square 12/6 15 volts A.C. 2½In. Flush circular moving iron 8/6 20 volts A.C. 3½in. Flush circular moving iron 25/- 300 volts A.C. 2½in. Flush circular moving iron 25/- 300 volts D.C. 2in. Flush square 300 volts D.C. 2in. Flush square			
50-0-50 amp. D.C. 2in. Flush square 12/6 15 volts A.C. 2µln. Flush circular moving 8/6 20 volts A.C. 3µlin. Flush circular moving 8/6 300 volts A.C. 2µlin. Flush circular moving 25/- 300 volts D.C. 2µin. Flush square 25/- 300 volts D.C. 2µin. Flush square 10 6			
15 volts A.C. 211n. Flush circular moving iron 8/6 20 volts A.C. 31in. Flush circular moving iron 25/- 300 volts A.C. 21in. Flush circular moving iron 25/- 300 volts D.C. 20in. Flush square 10 6			
20 volts A.C. 300 volts A.C. 300 volts A.C. 300 volts D.C. 21 iron iron 25/-			
20 volts A.C. 31 in. Flush circular moving iron 300 volts A.C. 300 volts D.C. 21 in. Flush circular moving iron 300 volts D.C. 21 in. Flush circular moving iron 	15 volts A.C.		
300 volts A.C. 25/- 300 volts A.C. 24in. Flush circular moving 300 volts D.C. 2in. Flush square			
300 volts A.C. 21/21. Flush circular moving iron 25/21. 300 volts D.C. 21/01. Flush square 10.6	20 volts A.C.		
300 volts D.C. 2in. Flush square 10 6	the second s	iron	
300 volts D.C. 2in. Flush square 10 6	300 volts A.C.		
		iron	
750 volts D.C. 2in. Flush square 106			
	750 volts D.C.	Zin. Hush square	10 6

100 MICROAMPS METERS

24in. circular flush mounting. Widely calibrated scale of 15 divisions marked "yards" which can be rewritten to suit requirements. These movements are almost unobtain-able today and being BRAND NEW IN MAKER'S CARTONS are a snip at ONLY 42/6.

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100-0-100 VOLTS METERS by San-gamo Weston. 2½in. circular, basic movement being 500-0-500 microamps A really first-class centre zero meter for hundreds of uses. BRAND NEW. IN MAKER'S CARTONS. ONLY 27/6.

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TRIPLETT UNIVERSAL TEST METER. Made by the famous Ameri-can meter manufacturers. Size 72in. x tilting bakelite container size 5100 meters, a unique tilting bakelite container size 5100 m $3\frac{1}{2}$ ohms per volt moving coil for D.C. measurements, and a first grade moving iron for A.C. Resistance readings up to 7.5 megohms, A.C. and D.C. volts to 1,000 D.C. current to 250 ma. In addition A.C. meter can be used as an output meter. Completely portable, with protective face cover. Thoroughly reconditioned. ONLY £9/19/6.

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CRYSTALS. British Standard 2-pin 500 kc/s. 15/-. Miniature 200 kc/s. and 465 kc/s. 10/- each.

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CHOKES. 10H 60 mA., 4/-, 5H 200 mA., 7/6, 20H 120 mA., 10/6 (post 1/- ca.).

MU-METAL SCREEN FOR VCR97 TUBE, etc., ONLY 8/6.

CABLE. CLEARANCE OFFER of 16:012 twin polythene. Weatherproof, and suitable for outdoor use, 39/6 per 100 yard coil (carriage, etc., 3/6). S.A.E. for sample, trade enquiries invited.

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NOVEMBER, 1955



Thousands of LAB Continuous Storage Units are daily solving the problem of control and storage of the great range of resistors. Compact, and capable of storing up to 720 separate resistors, LABpak make selection positive, simple and speedy. Now that Ceramicaps, Histabs and Wirewound resistors have been added to the carded range, the usefulness of LABpak storage units is enhanced.

FREE with any purchase of the LABpak range, these units are the complete answer to the storage problems of small production units, laboratories, etc.

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RECEIVER RI132A covers 100-124 Mc/s RECEIVER RIISZA covers 100-124 Mc/s with variable tuning. Ideal for conversion to F.M. Complete with II valves and circuit diagram. Brand new condition. Requires external power supply. Regret owing to limited number these are only available for callers. Price 65/-.

RECEIVER 25/73 (TRI196) superhet receiver with 465 kc/s I.F.s. Complete with all valves—2-EF39, I-EK32, 2-EF36, I-EBC33. BRAND NEW with full con-version data. Price 30/- plus 2/6 post.

R.F. UNITS. ALL BRAND NEW and BOXED. RF24, 20-30 Mc/s., 12/6. RF25, 40-50 Mc/s., 17/6. RF26, 50-65 Mc/s. 29/6 and RF27, 65-85 Mc/s., 32/6.

COMMUNICATIONS RECEIVER

RII55. BRAND NEW "MINT" con-dition, RII55A. £11/19/6. RII55B with super slow-motion drive, £12/10/-. RII55A, shop soiled models, as new, £9/19/6. Used models £7/19/6.

"L'and "N "MODELS. Covertrawler and shipping bands. Excellent condition £17/19/6. Carriage on all models 10/6 extra. All receivers supplied with FREE BOOKLET. Re-aligned and aerial-tested before despatch and gladly demon-strated to callers. Send S.A.E. for full details of power packs and receivers or 1/3 for booklet. DJF Loops and Visual Indicator Meters available.

A.C. MAINS POWER PACKS AND OUTPUT STAGE. All our power packs are guaranteed for 6 months.

Type A. In smart black metal case, size $8\frac{1}{2}$ in. $\times 4\frac{1}{2}$ in. $\times 6\frac{1}{2}$ in. less speaker, price $\frac{1}{24}$ 10/- plus 3/6 postage.

Type B. With built-in Sin. speaker in black metal case size $13\frac{1}{2}$ In. \times $5\frac{1}{2}$ In. \times 7 $\frac{1}{2}$ in. Price £5/5/- plus 5/- carriage.

Type C. With an Bin. speaker in specially designed beautiful black crackle cabinet to match receiver, size 11jin. × 10jin. × 6in. A de Luxe job. Price £6/10/- plus 5/- carriage. NOTE: 10/-REDUCTION WHEN PURCHASING ANY OF THE ABOVE POWER PACKS WITH RECEIVER.

45 Mc/s PYE STRIPS. Vision unit for London frequency. Complete with 6 EF60 and EA50. Circuit provided. Brand new. 59/6 each.

COMMUNICATIONS RECEIVER ADMIRALTY TYPE 8.28 (MARCONI CR.100). Valve line up 2 RF, F.C. separate local 05c., 3 I.F.S. 2nd Det., Output, B.F.O. and rectifier. Self contained power supply 200/250 volts A.C. 50 c/s. Variable selectivity (crystal filter). 6,000, 3,000, 1,200, 300 and 100 cycles. Frequency coverage 60 kc/s to 30 Mc/s in siz ranges, continuous except for rap, between 420 to 500 kc/s. Size 16in. x 194in. x 12in. Weight 82 Ib. The set for the serious operator. Thoroughly overhauled in superb condition, complete with new. valves and air tested prior to despatch. A real bargain at only £30.

E.M.I. OUTPUT METER

The instrument itself has basic movement of I mA and incorporates a fulland incorporates a full-wave instrument rectifier. There are two ranges 0-500 milliwatts and 0-5 watts and also a decibel scale. The Input impedance is 5,000 ohms. These are all BRAND NEW AND UNUSED in sealed Individual Manu-In sealed Individual Manufacturer's cartons, com-plete with instructions and are offered at about a third of list price. Only 35/- each plus 1/6 post.



R.C.A. AMPLIFIERS MODEL MI-11220



* Manufactured by Radio Corporation of America. × Brand new and unused.

- $\hat{\star}$ Power output 12 watts at 5-7.5-15-600Ω.
- Valve line up, 2 of 6L6, 4 of 6J7, 1 of 5U4. × *

Tapped transformer for use on 190-250 volts A.C. Grey crackled case, size 17×11×9in. *

GIRCUIT

SUPPLIED. ONLY £9. 19. 6 Plus 10/6 carriage (LESS VALVES, AS SHIPPED BY MAKERS) SET OF VALVES, NEW, BOXED, 59/6 POST FREE.

	METE	R B	ARGAINS		
RANGE	TYPE	SIZE			PRICE
50 Microamp.	D.C. M/C	21in.	flush circular		59/6
100 Microamp.	D.C. M/O	2110.	flush circular		39/6
500 Microamp.	D.C. M/C	2ín.	proj. (scaled 10	V.)	12/6
500-0-500 MicroA.	D.C. M/C	24in	fl. circ. (scaled	100-0-100	V) 25/0
] Milliamp.	D.C. M/C	2Hn.	flush circular		22/6
1 Milliamp.	D.C. M/C	211n.			25/-
5 Milliamp.	D.C. M/C	2in.			7/8
10 Milliamp.	D.C. M/O	2115.	fi. cire. (blank	scale)	10/6
100 Milliamp.	D.C. M/C	211n.	flush circular		10/6
150 Milliamp.	D.C. M/C	2ia.	dush square		7/6
1 Amp.	Thermo.	24in.	projection		6/9
3 Amp.	Thermo.	2Ĩn.	flush square		5/0
20 Amp.	D.C. M/O	2ln.	circular proj.		7/6
30-0-30 Amp.	D.C. M/1	Metal	cased. ('ar type		5/0
50-0-50 Amp.	D.C. M/C	2ln.	dush square		12/6
15 Volt	A.C. M/I	2110.	funh circular		8/6
300 Volt	A.C. M/I	2iln.	flush circular		25/-
300 Volt	D.C. M/C	2ĺn.	flush square		10/6

TRANSMITTER TYPE 12. Frequency coverage 1.2 to 17.5 Mc/s in four bands. M.O., Buffer, P.A., 2 stage Modulator, built-in stabilised A.C. mains power pack. P.A. Anode current meter M.C.W., C.W. and R.T. operation. V.F.O. or crystal controlled. Super "Table Topper" size 24in. x 12in. X 173in. Weight 1341b, Com-plete with all valves ready for operation. In first-class condition and tested before despatch with circuit and full Instructions. Only 22 plus £1 carriage. Can be demonstrated to caller.

GALVANOMETERS 79/6

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57, High Street, Hoddesdon, Herts. (Tel., Hodds. 3102.) [5244 A SSISTANTS (Scientific). The Civil Service A Commissioners invite applications for pen-sionable posts. Applications may be accepted up to 31st December, 1955, but early applica-tion is advised as an earlier closing date may be announced either for the competition as a whole or in one or more subject. Interview Boards will sit at frequent intervals. AGE at least 17¹⁰, and under 26 years of age on 1st January, 1955, with extension for regu-lar service in H.M. Forces, but candidates over 6 with specialised experience may be admitted. CANDIDATES must produce evidence of having reached a prescribed standard of education, particularly in a science or mathematical sub-ject. At least two years experience in the ducties of the class galned by service in the scientific rescention of the collowing or the Scientific subjects-(ii) CHENISTRY, bio-chemistry and metal-lurgy.

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Street, London, W.I. quoting No. S59/55. [S263] ELECTRONIC Engineers required for work on the application of radio valves for future development; work calls for vision and imagina-tion combined with circuit experience; O.N.C. or inter minimum qualifications; weekly or monthly staff vacancies available according to experience.—Quote EE/3, Personnel Dept., M.O. Vaive Co., Ltd., Brook Green, Ham-mersmith, w.6. [5104] DRAUGHTSMEN experienced in broadcast radio and television are offered exceptional opportunities with a firm of repute in the Lon-don area; experience in production drawing or mechanical design would qualify for a very attractive starting salary; applications, which will be trated in strict confidence, should give full details of exprisence and should baddressed to Box 6371. [NDIVIDUAL with real leadership and drive

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BANDI HIGH PASS FILTERS

CIRCUIT DIAGRAM

These filters reduce the effects of interference which may cause severe patterning on the screen. Although primarily designed for use with receivers embodying a 35 mc/s. I.F., they are also suitable for operation with all other I.F. frequencies.

Types available include C129, which is of minimum dimensions, and C 256, which gives an improved characteristic, for applications where a size increase can be accommodated.

We have also produced Cross-Over Units for the coupling of Band I and Band III aerials to a common feeder. Details are available on request.

ATTENUATION CHARACTERISTIC 60 FOR FILTER TYPE C. 129, FOR USE WITH 75 B FEEDER 50 90 IN TELEVISION RECEIVERS. ATTENUATION 40 30 20 10 60 50 FREQUENCY - MC/S. 40 0 20 30



THE TELEGRAPH CONDENSER CO. LTD SPECIAL PRODUCTS DIVISION North Acton London W3 Tel: ACOrn 0061

RECEIVER

FILTER

Wireless World

November, 1955

7 lb. REELS

Ersin Multicore 5-core Solder is supplied as standard for factory use in 9 gauges and 6 alloys on 7 lb reels. Prices on application.



RADIO & T/V SERVICE ENGINEERS 11b. REEL

For service engineers and others who use solder frequently, the 1 lb. pack is most econ-omical. It contains approximately 167 ft. of 18 s.w.g. 50 50 of 18 s.w.g. 50 50 alloy Ersin Multi-core 5-core Solder. Cat. Ref. R5018.15each (subject).

SIZE 1 CARIU	
This popular pack is	ERSIN
available containing	Blackie
Ersin Multicore 5-	STYLE .
core Solder in any	SOLDER
4 specifications.	
5/- each (subject)	State of the local division of the local div

	Catalogue Ref. No.	Alloy TinLead	s. W.G.	App. I'gth per carton
	C 16014	60/40	14	21 feet
	C 16018	60/40	18	55 feet
	C 14013	40/60	13	19 feet
	C 14016	40/60	16	38 feet

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HOME CONSTRUCTOR'S 2/6 PACK

This pack, con-taining 20 ft. of 18 s.w.g. 60/40 alloy Ersin Multicore Solder. wound on a reel, is just the right size for the home con-structor, Only 2/6, per carton (subject).



MULTICORE SOLD RS

SOLDER RINGS Butt jointed solder rings 1"-2" dia. and 10-22 s.w.g. are made from Ersin& Arax Multicore Solder at no extra cost for bulk quantities. Other sizes to special order O



(from 10 to 22 s.w.g.) supplied on 1 lb. and 7 lb. reels, Ersin Multicore 5-core Solder is also available in even gauges between 24 and 34 s.w.g. on 1 lh reels



descends. Three iron model PUBLICATIONS

Copies of the latest edition of Modern Solders-containing much useful information on melting points, standard gauges, constitution of alloys, fluxes, etc. - will gladly be sent to laboratory engineers and technicians applying on their firm's letterheading.

greater efficiency greater for speed

ARAX MULTICORE SOLDER

Recommended for metal fabrication. Solders almost all metals except alum-inium, In 3 alloys and 9 gauges on 7 lb. and 1 lb. reels and in and 6d. cartons



TAPE SOLDER

Only a match is needed to make a faultless joint with Multicore A MA Tape Solder. Cards 1/- each (subject). Also av-ailable for factory use in widths be-tween 1 and 2 and 1 and thickness from '005 on 31 b reels. Tape So 1451 - EGAS on 311b. reels

LIQUID & JELLY FLUX Ersin Flux is supplied in liquid form for dipping purposes. A dipping purposes. A high viscosity red jelly is also available for processes where a flux with good properties of adher-ence is required. It is approved by the Ministry of Supply for specific soldering purposes. Size 12 tins purposes. Size 12 tins 4/6 each (subject).



PRINTED CIRCUITS

Full details of the Multicore materials supplied for the efficient soldering of printed circuits is contained in publication P.C.101. Also included is information about Multicore Activated Surface Preservative.

greater economy on any soldering job

BIB WIRE STRIPPER AND CUTTER

This handy 3 in 1 tool strips insulation without nicking the wire, cuts wires cleanly and splits plastic extruded twin flex. Adjustable to most wire thic messes. Extraordinarily good value at 3 6 each (subject).



Recording tape can be jointed quickly and accurately. Incorp-orates many refinements usually only found on expensive models 18/6 each (subject)

BIB SOLDER THERMOMETER

Used for measuring bit or solder bath temperatures, this simple form of pyrometer will measure temperatures up () 400°C. The scale is graduated in degrees Centigrade and Faren-heit. Nett price £6.12 6.

SAVBIT TYPE 1 ALLOY

After extensive research and tests, Multicore have introduced Savbit Type 1 Alloy which prolongs the life of soldering iron bits by up to 10 times. Full details are available on application.

Five cores of extra-active Ersin Flux incorporated in Ersin Multicore Solder ensure that oxide deposits are quickly and efficiently cleaned from the surfaces to be soldered. The thin solder walls between the cores of flux melt quickly giving greater speed for precision soldering processes. Due to this increased speed of soldering, an alloy of lower tin content can often be used, resulting in considerable economies. Ersin Multicore 5-core Solder is approved by A.I.D., G.P.O. and A.R.B., complies with D.T.D.599 and R.C.S.1,000 and meets all pertinent U.S. Federal Specifications.

MULTICORE SOLDERS LTD., MULTICORE WORKS, HEMEL HEMPSTEAD, HERTS (BOXMOOR 3636)

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