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# Wireless World 

ELECTRONICS, RADIO, TELEVISION

## JUNE 1961

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VOLUME 67 No. 6.
PRICE: TWO SHILLINGS

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Iliffe Electrical Publications Ltd. Managing Director: H. S. Pocock, M.I.E.E. Dorset House, Stamford Street, London, S.E. 1

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This new manual of transistor circuitry has been prepared by Mullard engineers, as an up-to-date and readable volume which will be of use and interest to technicians, service engineers, junior designers and electronics students.
It has a page size of $8 \frac{1}{2}^{\prime \prime} \times 5 \frac{1}{2}^{\prime \prime}$ and describes more than 60 circuits-over 30 are made generally available for the first time-including both domestic and industrial applications.

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## Line Standards in 1946

IN recent weeks we have observed the growth of a myth which we feel should be dispersed before it becomes permanently entangled in the present gyratory arguments about television line standards. Its wide acceptance is understandable, for, if it is believed, it removes responsibility for the present line-standards dilemma (real or imagined) from the present generation and attributes it to the folly of our administrators in re-starting television too soon in 1946, and in any case on 405 lines!

What, then, should we have done? Waited for complete international agreement? We should still be waiting now. Tagged along, viewless and clueless, behind the rest of the field until a majority movement became apparent? Sweden began technical experiments with a 625 -line system in 1947 and was followed by Italy in 1949, but the first regular 625 television broadcasts (in Federal Germany) did not start until the end of 1950. Prior to that (in July 1950) a meeting of the Television Study Group of the International Radio Consultative Committee (C.C.I.R.) in Geneva, attended by delegates from Belgium, Denmark, Italy, the Netherlands, Sweden, Switzerland, France, the U.S.A. and the U.K., established standards which were recommended for those countries wishing to adopt a $625-\mathrm{line}$ system. These were endorsed together with three other systems on 405, 525 and 819 lines at the VIth Plenary Meeting of the C.C.I.R. in Geneva in the following year, but, and this point was made clear in our issue of August 1952, none of these systems was adopted as an international standard; nor has any been agreed upon to this day.

By the end of 1953 eleven countries in West Europe had either started broadcasting or were sending test transmissions on 625 lines, and Argentina, Turkey and Venezuela had also declared for this system. So we should have been pretty safe in starting a 625 -line service in that year or even in 1950 without much risk, but we should have lost four valuable years. Years in which television engineers and technicians, whose skill and knowledge had been extended by radar and communications work during the war would have been disbanded, and years in which this country would have lagged instead of leading in the post-war development of television.
All these factors were seen clearly by the Tele-
vision Committee appointed in 1943 under the chairmanship of Lord Hankey to prepare plans for the reinstatement and development of the television service after the war. This report emphasizes the need for an international standard, and because no one had yet thought of the magic number 625 the established 525 American standard was considered. They gave two reasons for rejecting it: that the improvement in picture quality over the 405 was not significant, and the differences in mains and field frequency would involve more serious problems than a mere change in the number of lines. Wisely they decided not to defer the reopening of the television service "for the uncertain period required to give an opportunity of incorporating some fundamental improvement in the system" but to restart broadcasting with 405 lines "which had achieved a high degree of reliability and afforded consistently good entertainment value in the home." The inherent limitations of the 405line system were squarely faced, and intensive research into a radically improved system with a line standard of the order of 1000 and with colour and stereoscopic effects was strongly urged to keep this country in the forefront of the television field While we are awaiting this or a comparable radical improvement we enjoy what is in practice the world's best television service.

## Belated Acknowledgment

WE apologize to our readers for what may have seemed a discourteous delay in thanking them for their congratulations and good wishes on the occasion of our Jubilee. It had been our intention to publish their letters, or at least some extracts from them in our June issue, but as they were still coming in from distant parts we delayed doing so. The file is now so voluminous that there is just not room in the journal to do justice to them all, and selection would, we feel, be invidious.

Our historical survey has evoked much reminiscence among our readers and the word nostalgia has been on everyone's lips (excepting "Free Grid's" who tells us that we have forgotten the original derivation and meaning of the word!). We claim his indulgence for the duration of these celebrations. They have been particularly pleasant for us, and we again thank our many friends for making them so.

Having, so to speak, crossed the line, we can now square away on our course with renewed confidence in the furure.

# Television Standards Conversion 

REMOVING THE 10C/s FLICKER IN 50c/s-60c/s FIELD CONVERSIONS

COONVERSION of television pictures from one standard to another has been common practice for a number of years, and has made possible the Eurovision link. However, conversion of European standards, which all employ a $50 \mathrm{c} / \mathrm{s}$ field frequency, to the 525 -line $60 \mathrm{c} / \mathrm{s}$ system in use in North America, and vice versa, has set a new problem. In addition to the relatively easily removed moiré pattern, caused by the interaction of closely related linestructures, a new type of picture deformation becomes apparent. The different field frequencies interact one with the other to produce a $10 \mathrm{c} / \mathrm{s}$ amplitude modulation of the video signal, which appears as a flicker at this frequency, and which renders the converted picture intolerable to the viewer.
Equipment designed to overcome this problem has been developed by the B.B.C. Engineering Division and has already been used to handle recordings, for translation to the American standard, of the Paris Summit Conference and the wedding of H.R.H. Princess Margaret.

The principle of a standards converter is shown in diagrammatic form in Fig. 1. The incoming video signal is amplified and displayed on a high-quality cathode-ray tube which has an afterglow time constant of 7 msec . A lens focuses the displayed picture on to the storage mosaic of a camera tube scanned at the rate corresponding to the new standard, the output of which is processed in a sync-pulse insertion and blanking unit. The signal emerging from this unit is the converted signal.

As has been mentioned, when the number of lines per field of the two standards under consideration are closely related the picture on the new standard is subject to moire interference caused by the fact that the lines of the two systems will be at very slightly different angles to the horizontal. In practice this effect is removed by the addition of "spot wobble" to the displayed incoming picture of such an amplitude as to camouflage the line structure of each field.

The mechanism producing the $10 \mathrm{c} / \mathrm{s}$ flicker is indicated in Fig. 2. The top curve is a graph of the brightness of a single displayed picture element plotted against time and it is assumed that the signal at this point is peak white. It will be seen that the
brightness increases to peak white as the element is scanned, once every 16.6 msec , and that it then dies away exponentially. The time constant of this decay is, with the cathode-ray tube used, 7 msec . The middle curve is the charge on the C.P.S. Emitron camera tube target, which in this case is being scanned at a $50 \mathrm{c} / \mathrm{s}$ field frequency. The charge would increase linearly with time if display brightness were constant, but as the brightness of the display tube decays exponentially the charging curve of the target is also exponential.
The target charge increases in this way until the display tube element is rescanned. The camera tube element has not yet been neutralized by its scanning beam and continues to increase for a further 3.4 msec , when it is discharged. It now commences to charge again, but at a lower rate, as the display tube element brightness has decayed from its peak level. The action continues in this manner until the fifth camera tube scan, when the two scans are once more coincident. If the peaks of the target charging curve are joined, a sawtooth type of waveform is obtained at a frequency of $10 \mathrm{c} / \mathrm{s}$, and this is proportional to the output of the camera tube. The signal is thereby amplitude modulated at $10 \mathrm{c} / \mathrm{s}$.

Removal of this flicker is effected by a pulsesampling, manually adjusted amplitude compensator, followed by an automatic compensator which removes any flicker remaining after the signal has been subjected to the effect of the manually set circuit. The system, which is the subject of British Patent 801140 (T. C. Macnamara and A. M. Spooner), operates in the manner indicated in Fig. 3. The incoming video signal is passed through a lowpass filter which is switch-selected to pass the band of frequencies used by the incoming standard ( $3 \mathrm{Mc} / \mathrm{s}$ for $405-$ line $50 \mathrm{c} / \mathrm{s}$ standard). The next process is the insertion at the end of each line of the picture of a reference pulse which has a width of about $2 \%$ of the line period, and an amplitude of about $85 \%$ peak white. The pulse slightly overlaps the right-hand edge of a line of video signal and is apparent as a bright vertical stripe on the picture. The video signal with the reference pulse added is now amplified and displayed on the cathoderay tube. The conversion process referred to earlier


Fig. 1. The basic principle of a standards converter.


Fig. 2. The mechanism producing the $10 \mathrm{c} / \mathrm{s}$ amplitude modulation of the video signal, in a conversion from 525 lines 60 fields $/ \mathrm{sec}$ to 405 lines 50 fields $/ \mathrm{sec}$.

Fig. 3. Block diagram of B.B.C. $50 \mathrm{c} / \mathrm{s}-60 \mathrm{c} / \mathrm{s}$ converter, showing units concerned with $10 \mathrm{c} / \mathrm{s}$ flicker removal. Woveforms shown are of peak white lines.



The B. B.C. standards converter. From left to right: the flicker correction bay, the camera and display control unit with monitor, and the display pedestal including the comera and display tube.
this beng used to drive a modulator which is in series with the signal path. The manual gain control adjusts the amplitude of the correction signal until the flicker is at minimum level. After this compensator the signal passes through a similar circuit which functions as an a.g.c. After passing through a modulator the signal feeds the sampling gate, which then applies the derived correction signal to the series modulator. The whole circuit is a feedback loop and is capable of reducing amplitude variations at
$10 \mathrm{c} / \mathrm{s}$ from $30 \%$ to less than $1 \%$. The signal, then, first passes through a manually adjusted compensator, wath reduces the level of modulation to a value which is within the range of control of the automatic circuit.

On leaving the two compensating circuits the signal is processed in a unit which inserts sync pulses and blanking intervals, and in which the reference pulses are removed. The signal is now converted to the new standard and ready for use.

## BOOKS RECEIVED

A Textbook of Electricity, by H. G. Mitchell. A school textbook suitable for use up to advanced level G.C.E., and by University Scholarship candidates. A complete account of electric and magnetic phenomena is given, and the rationalized m.k.s. system of units is used throughout. A short description of technological applications is included, and the final chapters include discussions of atomic and nuclear physics, the quantum theory and a physical picture of wave-mechanics. Some 300 questions, taken from recent examination papers are set. Pp. 598; Illustrated. Methuen \& Co., Ltd., 36, Essex Street, London, W.C.2. Price 25 s.
V.H.F. Line Techniques, by C. S. Gledhill. Designed for both undergraduate and practising engineer, the book seeks to impress the advantages of the circlediagram over the more traditional, analvtical methods of transmission-line problem solution. Worked examples of typical problems are given, which enable the work to be used as a reference book. Pp. 60; Figs. 51. Edward Arnold (Publishers), Ltd., 41, Maddox Street, London, W.1. Price 12 s 6 d .

How To Use Meters, by John F. Rider and Sol D. Prensky. The second edition of a practical handbook on the principles and operation of all types of meters.

The basic types of movement are described, and their use as voltage and current meters, with descriptions of more sophisticated instruments such as chopper electrometers and valve-voltmeters. Information is given on measurement of quantities in receivers and transmitters and a chapter is included on more sophisticated electronic measurements. Pp. 210; Figs. 181. John F. Rider Publisher, Inc., 116, West 14th Street, New York 11, N.Y. Price $\$ 3.90$.

Hi-Fi Amplifier Circuits, by E. Rodenhuis, translated by G. Du Cloux. For all who are interested in the design and construction of high-quality audio amplifiers. A chapter on general design cons:derations is followed by descriptions of several power amplifiers and preamplifiers, including a four-channel mixer. Pp. 105; Figs. 64. Philips' Technical Library. Obtainable from Cleaver Hume Press, Ltd., 31, Wright's Lane, Kensington, London, W.8. Price 15 s .

Boolean Algebra and its Applications, by J. E. Whitesitt. Introduction to algebra of logical deductions with particular reference to computer and switching circuit applications. Pp. 182; Figs. 100. Addison-Wesley Publishing Co., Inc., 10-15, Chitty Street, London, W.1.
Price 51s.

DEFINITIONS AND INTER-RELATIONSHIPS

ANYONE who has glanced at the manufacturer's figures for transistor characteristics must have been bewildered, as I was, by their profusion. Figures are quoted for some or all of about thirty different quantities, with careful stipulations as to which of the three configurations (grounded base, grounded emitter or grounded collector) these figures relate to.
How easy-going seem those far-off days of thermionic valves, when only three figures appeared, namely $g_{m}, R_{a}$, and $\mu$-and anyone who pursued a course at a Technical College would have learned at a very early stage that $\mu$ equals $g_{m} \times \mathrm{R}_{a}$, so only two parameters were really needed!
It is the purpose of this article to sort out and relate the transistor parameters, beginning with a description of the transistor in terms of its static characteristics, and including one or two digressions which may be of interest.
Transistor Behaviour.-The simple transistor is basically a three-terminal active network. Very briefly its action is as follows: for suitable voltages between the electrodes, the current flowing into the emitter divides between the collector and base, with about $97 \%$ of it flowing to the collector and about $3 \%$ flowing to the base.

Variations of the voltages produce corresponding variations of these currents, and whilst the overall


Fig. 1. Three voltage and three current variables for the transistor.
currents remain roughly in the above proportion, the incremental changes of each current depend on the incremental changes of voltage.
It is true to say, however, that a change of emitter current produces a roughly comparable change of collector current, whilst a change of base current is accompanied by a much larger change in collector current. For this reason, the base current is treated as an independent variable.
We may compare the above behaviour with that of a triode in which the anode current is slightly affected by the anode voltage, but very much more affected by the grid voltage, and so the grid voltage is usually treated as an independent variable.

We have six variables-three voltages and three currents-as shown in Fig. 1. Two suffixes are used for each voltage, their order defining which electrode is positive with respect to the other.
Two fundamental equations hold:-

$$
\begin{aligned}
\mathrm{V}_{e c} & =\mathrm{V}_{e b}+\mathrm{V}_{b c} \text { and } \\
\mathrm{I}_{e} & =\mathrm{I}_{b}+\mathrm{I}_{c}
\end{aligned}
$$

so in fact there are only four variables, which four being entirely a matter of choice.

Fig. 2. Groundedemitter configuration for the transistor.


If, out of these four, any two are fixed externally, the transistor fixes the other two. Thus if a certain voltage is applied between emitter and base, and another voltage between base and collector, the emitter-collector voltage is determined by the first equation above, and the three currents $I_{e}, I_{b}$ and $I_{c}$ are determined by the transistor and the second equation.

It is equally true to say, if a given emitter-collector voltage is applied, in order that a given base current shall flow, a unique value of emitter-base voltage will be needed, and so unique values of collector current and emitter current will flow. Once again, therefore, two variables, in this case $\mathrm{V}_{e c}$ and $\mathrm{I}_{b}$, effectively determine the other four.

It is important to realize that, whilst the voltages are in a sense the "prime movers" and canse the currents to flow, this does not mean that we cannot regard a current as an independent variable -in fact, as we saw above, the base current is usually so considered.

Static Characteristics of a Transistor.-Since we have four separate variables, two of which are independent and two dependent, the characteristics of a transistor cannot be expressed by means of one set of curves, as can those of a valve. Two sets are needed, each incorporating either one voltage and two currents or two voltages and one current, the two sets together covering two currents and two voltages. Evidently a set incorporating the three voltages or the three currents is useless, as these are related by the fundamental equations anyway.
Which variables to plot against which is once

[^0]

Fig. 3. Typical output curves for a transistor in the grounded-emitter configuration.
again a matter of convenience. The following table gives a few of the possibilities:-
the first four sets of curves respectively in the above table.

Typical output curves are shown in Fig. 3. They closely resemble those of a pentode valve, except that the parameter $I_{b}$ takes the place of the corresponding parameter $\mathrm{V}_{g}$.

By what has been said above, analogous characteristics for either of the other two configurations (grounded base and grounded collector-see Figs. 4a and b) may be derived from these.

Transistor Parameters.-Three quite separate sets of parameters are defined expressing the behaviour of a transistor tor small signal amplification. These are: (i) The Hybrid, or " $h$ ", parameters; (ii) The T -network parameters; and (iii) The Mullard parameters.
Each of these sets is in turn subdivided into three sets-undashed, dashed and double-dashed-respectively expressing the small-signal behaviour of the transistor in the grounded-base, groundedemitter and grounded-collector configuration.
Three other quantities $a, a^{\prime}$, and $a^{\prime \prime}$ are further defined, rather loosely called "current amplification" ( $\mathrm{dI}_{\text {out }} / \mathrm{dI}_{\text {in }}$ ), for the three configurations.

Hybrid Parameters.-To define the $h$-parameters, the transistor is regarded as an active four-terminal

| $\mathbf{X}$ axis (first independent variable) | .. | .. | $\ldots$ | $\ldots$ | $\mathrm{V}_{e c}$ | $\mathrm{~V}_{e c}$ | $\mathrm{I}_{b}$ | $\mathrm{I}_{b}$ | $\mathrm{~V}_{b c}$ | $\mathrm{~V}_{b c}$ | $\ldots$ |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Parameter (second independent variable) | $\ldots$ | $\ldots$ | $\ldots$ | $\mathbf{I}_{b}$ | $\mathrm{I}_{b}$ | $\mathrm{~V}_{e c}$ | $\mathrm{~V}_{e c}$ | $\mathrm{I}_{b}$ | $\mathrm{I}_{b}$ | $\ldots$ |  |  |
| Y axis (dependent variable) | $\ldots$ | $\ldots$ | .. | $\ldots$ | $\ldots$ | $\mathrm{I}_{c}$ | $\mathrm{~V}_{e b}$ | $\mathrm{I}_{c}$ | $\mathrm{I}_{e b}$ | $\mathrm{I}_{e}$ | $\mathrm{~V}_{e b}$ | $\ldots$ |

There are 108 permissible sets in all. In those given here, $I_{b}$ has been chosen as an independent variable for the reason given above. A bracketed pair supplies the whole information about the transistor's behaviour and consequently, given one such pair, it is possible to plot from it any of the remaining 106 sets of curves.

It is possible to choose the variables in such a way that both the sets needed may be plotted on one sheet of graph paper, for example:-

| $\mathbf{X}$ axis | $\cdot$ | $\cdots$ | $\mathbf{I}_{c}$ | $\mathbf{I}_{c}$ |
| :--- | :--- | :--- | :--- | :--- |
| Parameter | .. | .. | $\mathbf{I}_{b}$ | $\mathbf{I}_{b}$ |
| $\mathbf{Y}$ axis | .. | .. | $\mathrm{V}_{e c}$ | $\mathrm{~V}_{e b}$ |

but although the first of these is a re-arrangement of the "output characteristic" in the groundedemitter configuration, the second set is of little practical value.
Four sets of curves are usually supplied though only two are necessary. Since the transistor is usually used in the grounded-emitter configuration as in Fig. 2, the four sets are plotted for this configuration. For the reason set out above, $I_{b}$ is treated as an independent variable in each set. Also, to enable the location of a working point, one of the sets relates the output current ( y axis) with the output voltage ( $x$ axis). The four sets are called "output", "voltage feedback", "current transfer" and "input" characteristics, and are
network, the lower pair of terminals being connected together and joined to the "common" electrode, i.e., base, emitter or collector (see Fig. 5). In this diagram $\mathrm{I}_{1}$ signifies the input current, $\mathrm{I}_{2}$ the output current, $\mathrm{V}_{1}$ the input voltage and $\mathrm{V}_{2}$ the output voltage.
Fixing any two of these quantities automatically


Fig. 4. (a) Grounded-base and (b) grounded-collector configuration for the transistor.


Fig. 5. Four-terminal network treatment of a transistor.


Fig. 6. Four-terminal network treatment of a transistor in the grounded-base configuration.

(a)
(b)


Fig. 7. Two four-terminal networks connected as in (a) con be reduced to a single four-terminal network as in (b) whose hybrid parameters ore each the sum of the two corresponding hybrid porometers of the two fourterminal networks of (o).
fixes the other two, as was explained above. Thus if we take $I_{1}$ and $V_{2}$ as independent variables, we may express the dependent variables $\mathrm{V}_{1}$ and $\mathrm{I}_{2}$ in terms of these as follows:-

$$
\begin{aligned}
& d V_{1}=\left(\frac{\delta V_{1}}{\delta I_{1}}\right)_{V_{2}} d I_{1}+\left(\frac{\delta V_{1}}{\delta V_{2}}\right)_{I_{1}} \mathrm{dV}_{2} \text { and } \\
& \mathrm{dI}_{2}=\left(\frac{\delta I_{2}}{\delta I_{2}}\right)_{V_{2}} d I_{1}+\left(\frac{\delta I_{2}}{\delta V_{2}}\right) \mathrm{I}_{1} \mathrm{dV}_{2}
\end{aligned}
$$

For small increments, the partial differential coefficients may be considered constant, so the equations may be re-written -

$$
\begin{aligned}
& \mathrm{dV}_{1}=h_{11} \mathrm{dI}_{1}-h_{12} \mathrm{dV}_{2} \text { and } \\
& \mathrm{dI}_{2}=-h_{21} \mathrm{dI}_{1}+h_{22} \mathrm{dV}_{2}
\end{aligned}
$$

where $h_{11}, h_{12}, h_{21}$ and $h_{22}$ are the "hybrid parameters". The name arises from the fact that $h_{11}$ has dimensions of resistance, $h_{22}$ has dimensions of conductance, while $h_{12}$ and $h_{21}$ are pure numbers. The negative signs are conventional.

Actually these symbols represent the parameters for the transistor in the grounded-base configuration (see Fig. 6), and it is easily scen, by considering the action of a transistor, that $h_{11}$ is positive, $h_{12}$ is also positive since $\left(\frac{\delta \mathrm{V}_{1}}{\delta \mathrm{~V}_{2}}\right)_{\mathrm{I}_{1}}$ is negative, $h_{21}$ is negative since $\left(\frac{\delta \mathrm{I}_{2}}{\delta \mathrm{I}_{1}}\right)_{\mathrm{V}_{2}}$ is positive, and $h_{22}$ is positive.


(c)

(d)

Fig. 8. (a), (b) and (c) Three configurations of a transistor and (d) four-terminal network treatment of a transistor.

Corresponding $h$-parameters for the groundedemitter and grounded-collector configurations are defined by exactly the same equations, but the " $h$ "s are dashed and double-dashed respectively, and have correspondingly different values.

An interesting property of hybrid parameters follows from their definition. If two four-terminal networks, having sets of parameters $\mathrm{H}_{1}$ and $\mathrm{H}_{2}$, are connected as shown in Fig. 7(a), they can be reduced to a single four-terminal network whose four parameters, forming a set H , are the sum of the corresponding parameters of sets $\mathrm{H}_{1}$ and $\mathrm{H}_{2}$, as shown in Fig. 7(b).

Sinusoidal Voltages and Currents.-The equations may be re-written:-

$$
\begin{aligned}
& \mathrm{V}_{1}=h_{11} \mathrm{I}_{1}-h_{12} \mathrm{~V}_{2} \text { and } \\
& \mathrm{I}_{2}=-h_{21} \mathrm{I}_{1}+h_{22} \mathrm{~V}_{2}
\end{aligned}
$$

where $V_{1}, I_{1}, V_{2}$, and $I_{2}$ are complex numbers (" rotating vectors ") representing sinusoids of small amplitude.

For low frequencies, the coefficients $h_{11}$, etc., are real numbers. For higher frequencies, it is necessary to regard them as complex functions of frequency.

Relation Between $h$-Parameters for Different Configurations.-Figs. 8(a), (b) and (c) show the transistor connected in its three possible configurations, along with applied voltages and currents, while Fig. 8(d) shows the transistor treated as a four-terminal network with its input and output voltages and currents.

Comparing the diagrams $8(\mathrm{a})$, (b) and (c) in turn with (d), we have, for the grounded-base configura-tion:-

$$
V_{e b}=V_{1}, I_{e}=I_{1}, \quad I_{c}=I_{2} \text { and } V_{b c}=V_{2}
$$

and therefore

$$
\mathrm{dV} e b=h_{11} \mathrm{~d} \mathrm{I}_{e}-h_{12} \mathrm{dV}_{b c} \text { and }
$$

$$
\mathrm{d} \mathrm{I}_{c}=-\hat{h}_{21} \mathrm{~d} \mathbf{I}_{e}+h_{2 \underline{2}} \mathrm{dV}_{b \mathrm{c}}
$$

For the grounded-emitter configuration we have:$\mathrm{V}_{e b}=-\mathrm{V}_{1}, \mathrm{I}_{b}=-\mathrm{I}_{1}, \mathrm{I}_{c}=\mathrm{I}_{2}$, and $\mathrm{V}_{e c}=\mathrm{V}_{2}$ and therefore

$$
\overline{\mathrm{dV}}{ }_{e b}=-h^{\prime}{ }_{11} \mathrm{dI}_{b}-h_{{ }_{12}^{\prime}} \mathrm{dV}_{e c} \text { and }
$$

For the grounded-collector configuration we have:-

$$
\mathrm{V}_{b c}=\mathrm{V}_{1}, \mathrm{I}_{b}=-\mathrm{I}_{1}, \mathrm{I}_{e}=-\mathrm{I}_{2} \text {, and } \mathrm{V}_{e c}=-\mathrm{V}_{2}
$$ and therefore

$$
\begin{aligned}
& \mathrm{dV}_{b e}=-h^{\prime \prime}{ }_{11} \mathrm{dI}_{b}+h^{\prime \prime}{ }_{12} \mathrm{dV}_{e c} \text { and } \\
& -\mathrm{dI}_{e}=h^{\prime \prime}{ }_{21} \mathrm{dI}_{b}-h^{\prime \prime}{ }_{22} \mathrm{dV}_{e c}
\end{aligned}
$$

We also have the two Kirchhoff equations

$$
\begin{aligned}
& \mathrm{V}_{e c}=\mathrm{V}_{e b}+\mathrm{V}_{b c} \text { and } \\
& \mathrm{I}_{e}=\mathrm{I}_{b}+\mathrm{I}_{c}
\end{aligned}
$$

and we can therefore find all the members of any one set of parameters in terms of those of any other set.

We will derive, as a demonstration, $h_{21}^{\prime}$ and $h_{22}^{\prime}$ in terms of $h$-parameters, mercly quoting the remainder.

$$
\begin{aligned}
& \mathrm{dV}_{e b}=h_{11} \mathrm{dI}_{e}-h_{12} \mathrm{dV}_{b c} \\
& =h_{11} \mathrm{dI}_{b}+h_{11} \mathrm{dI}_{c}-h_{12} \mathrm{dV}_{e c}+h_{12} \mathrm{dV}_{e b} \\
& =h_{11} \mathrm{dI}_{l} /\left(1-h_{12}\right)+h_{11} \mathrm{dI}_{c} /\left(1-h_{12}\right)- \\
& h_{12} \mathrm{dV}_{e c} /\left(1-h_{12}\right) \\
& \mathrm{dI}_{c}=-h_{21} \mathrm{dI}_{e}+h_{22} \mathrm{dV}_{b c} \\
& =-h_{21} \mathrm{dI}_{b}-h_{21} \mathrm{dI}_{c}+h_{22} \mathrm{dV}_{e c}- \\
& =-h_{21} \mathrm{dI}_{b} /\left(1+h_{21}\right)+ \\
& h_{22} \mathrm{dV}_{e c} /\left(1+h_{21}\right)-h_{22} \mathrm{dV}_{e b} /\left(1+h_{21}\right)
\end{aligned}
$$

Therefore

$$
\begin{aligned}
& \mathrm{dI}_{c}=-h_{21} \mathrm{dI}_{b}-h_{21} \mathrm{dI}_{c}+h_{22} \mathrm{dV}_{e c}- \\
& h_{11} h_{22} \mathrm{dI}_{b} /\left(1-h_{12}\right)-h_{11} h_{22} \mathrm{dI}_{c} /\left(1-h_{12}\right) \\
&+h_{12} h_{22} \mathrm{dV}_{e c} /\left(1-h_{12}\right) \\
&=\left[\frac{-h_{21}-h_{11} h_{22} /\left(1-h_{12}\right)}{1+h_{21}+h_{11} h_{22} /\left(1-h_{12}\right)}\right] \mathrm{dI}_{b} \\
&+\left[\frac{h_{22}+h_{12} h_{22} /\left(1-h_{12}\right)}{1+h_{21}+h_{11} h_{22} /\left(1-h_{12}\right)}\right] \mathrm{dV}_{e \sigma}
\end{aligned}
$$

Comparing this with

$$
\mathrm{dI}_{c}=h^{\prime}{ }_{21} \mathrm{dI}_{b}+h_{22}^{\prime} \mathrm{dV}_{e c}
$$

we see that

$$
h_{21}^{\prime}=\left(h_{12} h_{21}-h_{11} h_{22}-h_{21}\right) / \mathrm{D} \text { and } h_{22}^{\prime}=h_{22} / \mathrm{D}
$$

where $\mathrm{D}=\left(1+h_{21}\right)\left(1-h_{12}\right)+h_{11} h_{22}$
Summarizing the results, we have:-

$$
\begin{aligned}
& h_{11}^{\prime}=h_{11} \mathrm{D}, h_{22}^{\prime}=h_{22} / \mathrm{D}, h_{12}^{\prime}=1-\left(1+h_{21}\right) / \mathrm{D} \\
& \text { and } h_{21}^{\prime}=\left(1-h_{12}\right) / \mathrm{D}-1 \\
& h_{11}^{\prime \prime}=h_{11} / \mathrm{D}, h_{22}^{\prime \prime}=h_{22} / \mathrm{D}, h_{12}^{\prime \prime}=\left(1+h_{21}\right) / \mathrm{D} \\
& \text { and } h^{\prime \prime}{ }_{21}=-\left(1-h_{12}\right) / \mathrm{D}
\end{aligned}
$$

We can write down certain simple relations, for example:-

$$
\begin{aligned}
& h_{11}^{\prime}=h^{\prime \prime}{ }_{11}, h_{22}^{\prime}=h^{\prime \prime}{ }_{22} h_{12}^{\prime}+h_{12}^{\prime \prime}=1, \\
& \left.h_{{ }_{31}}^{\prime}+h^{\prime \prime}{ }_{21}=1 \text { and }^{\prime \prime} h_{11}^{\prime} h_{22}^{\prime}-h_{12}^{\prime} h_{21}^{\prime}\right)= \\
& \left(h_{11} h_{22}-h_{12} h_{21}\right)\left(h_{11}^{\prime \prime} h_{22}^{\prime \prime}-h_{21}^{\prime \prime} h_{12}^{\prime \prime}\right)
\end{aligned}
$$

Measurement of $h$-Parameters.-From the equations

$$
\begin{aligned}
& \mathrm{dV}_{e b}=h_{11}^{\prime} \mathrm{dI}_{b}+h_{1}^{\prime} \mathrm{d}^{\prime} \mathrm{dV}_{e c} \text { and } \\
& \mathrm{dI}_{c}=h_{21}^{\prime} \mathrm{dI}_{b}+h^{22} \mathrm{dV}_{e c}
\end{aligned}
$$

we have
$h_{\mathrm{II}}^{\prime}=\left(\frac{\delta \mathrm{V}_{e b}}{\delta \mathrm{I}_{b}}\right), \mathbf{V}_{e c}$ which is the slope of the input characteristic,


Fig. 9. T-network representation of a transistor.


Fig. 10. T-network representation of a transistor in the grounded-emitter configuration.

$$
h_{12}^{\prime}=\left(\frac{\delta \mathrm{V}_{e b}}{\delta \mathrm{~V}_{e c}}\right), \text { which is the slope of the }
$$ voltage feedback characteristic,

$$
h^{\prime}{ }_{21}=\left(\frac{\delta I_{c}}{\delta \mathrm{I}_{b}}\right),_{V_{e c}} \text { which is the slope of the }
$$ current transfer characteristic, and

$$
h_{22}^{\prime}=\left(\frac{\delta \mathrm{I}_{c}}{\delta \mathrm{~V}_{e c}}\right), \text { which is the slope of the }
$$ output characteristic.

All of these are in fact positive, as can be seen by considering the transistor action.

We have to calculate the values of the groundedbase and grounded-collector parameters from these slopes, using the relations quoted above.

One very important point must be noted: the relations calculated above hold only for increments about the same working point in the three configurations. Thus, if the values of $h$ and $h^{\prime}$ are quoted for $\mathrm{V}_{\mathrm{c}}=-2 \mathrm{~V}$ and $\mathrm{I}_{c}=1 \mathrm{~mA}$, we must remember that " ${ }^{c} \mathrm{~V}_{c}=-2 \mathrm{~V}$ " in the grounded-base configuration means " $\mathrm{V}_{b c}=2 \mathrm{~V}$ ", whilst " $\mathrm{V}_{\mathrm{c}}=-2 \mathrm{~V}$ " in the grounded-emitter configuration means " $\mathrm{V}_{e c}=$ $2 V^{\prime \prime}$, so in fact the values of $h$ and $h^{\prime}$ are quoted for slightly different working points since $\mathrm{V}_{\text {ec }}$ usually exceeds $\mathrm{V}_{b c}$ by about 0.2 V .

Hybrid parameters must be regarded as the fundamental parameters describing the behaviour of any four-terminal network (active or passive, linear or non-linear, frequency sensitive or otherwise). They are somewhat too academic, however, for an clectronic engineer seeking to analyze transistor circuits.

T-Network Parameters.-When considering the behaviour of a transistor in circuit, it is convenient to represent it as a T-network consisting of three resistances and one active element, as shown in Fig. 9.

The four parameters (clearly four, and only four,


Fig. 11. T-network representation of a transistor in the grounded-collector configuration.
separate constants will be required whatever system of pirameters is adopted) in this case are $r_{1}, r_{2}$, $r_{3}$ and $r_{4}$.

The equations relating $V_{1}$ with $I_{1}$ and $V_{2}$, and $I_{2}$ with $I_{1}$ and $V_{2}$, are respectively:-

$$
\begin{aligned}
& \mathrm{V}_{1}=\left(r_{1} r_{2}+r_{1} r_{3}+r_{8} r_{3}-r_{3} r_{4}\right) \mathrm{I}_{1} /\left(r_{2}+r_{3}\right)- \\
& r_{3} \mathrm{~V}_{2} /\left(r_{2}+r_{3}\right) \mathrm{I}_{3}\left(r_{3}\right) \\
& \mathrm{I}_{2}=\left(r_{3}+r_{4}\right) \mathrm{I}_{1} /\left(r_{2}+r_{3}\right)+\mathrm{V}_{2} /\left(r_{2}+r_{3}\right)
\end{aligned}
$$

By comparing these equations with those relating the same quantities in terms of $h$-parameters, we obtain the following relations between the resistances in this representation and the $h$-parameters:-

$$
\begin{aligned}
& h_{11}=r_{1}+r_{3}\left(r_{2}-r_{4}\right) /\left(r_{2}+r_{3}\right), h_{12}=r_{3} /\left(r_{2}+\right. \\
& \left.r_{3}\right), h_{21}=-\left(r_{3}+r_{4}\right) /\left(r_{2}+r_{3}\right) \text { and } h_{22}=1 / \\
& \left(r_{2}+r_{3}\right)
\end{aligned}
$$

For the grounded-base configuration, terminal 1 is the emitter and $r_{1}$ is called the emitter resistance $r_{e}$; terminal 2 is the collector and $r_{2}$ is called the collector resistance $r_{c}$, and terminals 3 and 4 are the base and $r_{3}$ is called the base resistance $r_{b}$. Resistance $r_{4}$ is usually denoted $r_{m}$. So, for this configuration the above relations yield:-

$$
\begin{aligned}
& r_{b}=r_{3}=h_{12} / h_{22}, r_{c}=r_{2}=\left(1-h_{12}\right) / h_{22}, r_{e}= \\
& r_{1}=\left(h_{11} h_{22}-h_{12} h_{21}-h_{12}\right) / h_{22}, \text { and } r_{m}=r_{4}= \\
& -\left(h_{12}+h_{21}\right) / h_{22}
\end{aligned}
$$

For the grounded-emitter configuration similar equations apply, except that the coefficients of $I_{1}$ and $\mathrm{V}_{2}$ are dashed, yielding therefore:-

$$
\begin{aligned}
& r_{1}=\left(h_{11}^{\prime} h_{22}^{\prime}-h_{12}^{\prime} h_{21}^{\prime}-h_{12}^{\prime}\right) / h_{22}^{\prime} \\
& r_{4}=-\left(h_{12}^{\prime}+h_{21}^{\prime}\right) / h_{22}^{\prime}
\end{aligned}
$$

and
Substituting $h$-parameters for $h^{\prime}$ ones and using the relations derived for $r_{b}, r_{c}, r_{e}$ and $r_{m}$, we obtain: $r_{1}=r_{b}, r_{2}=r_{0}-r_{m}, r_{3}=r_{g}$ and $r_{4}=-r_{m}$ giving the T -network representation as shown in Fig. 10.


Above: Fig. 12. Four passive elements as shown cannot fully represent a transistor.

For the grounded-collector configuration, the coefficients of $I_{1}$ and $V_{2}$ in the equations are doubledashed, and so

$$
\begin{aligned}
& r_{1}=\left(h_{11}^{\prime \prime} h_{22}^{\prime \prime}-h_{12}^{\prime \prime} h_{21}^{\prime \prime}-h_{12}^{\prime \prime}\right) / h_{22}^{\prime \prime} \ldots \text { and } \\
& r_{4}=-\left(h_{12}^{\prime \prime}+h_{21}^{\prime}\right) / h_{22}^{\prime \prime}
\end{aligned}
$$

Again, substituting $h$-parameters for $h^{\prime \prime}$ ones and using the relations above for $r_{b}, r_{c}, r_{e}$ and $r_{m}$, we get:-

$$
r_{1}=r_{b}+r_{m}, \mathrm{r}_{2}=r_{e}, r_{3}=r_{c}-r_{m} \text { and } r_{4}=r_{m}
$$ so the equivalent T-network becomes as shown in Fig. 11.

One advantage of the T-network representation is immediately apparent-the elements of the equivalent circuits for the grounded-emitter and grounded-collector configurations are extremely simply related to those for the grounded-base configuration.
Other Equivalent Circuits.-The representarion above included three passive and one active element. Four separate constants are required; one therefore asks oneself if a representation such as Fig. 12, containing four separate passive elements, will do. If however one forms equations for this network in the form $\mathrm{V}_{1}=h_{11} \mathrm{I}_{1}-h_{12} \mathrm{~V}_{2}$ and $\mathrm{I}_{2}=-h_{21} \mathrm{I}_{1}+$ $h_{22} \mathrm{~V}_{2}$, where the $h$ coefficients are functions of the four resistances, one finds that $h_{21}=h_{12}$. This representation therefore only provides us with three separate $h$-parameters and cannot therefore represent a transistor. We could have predicted this from the theorem of reciprocity; a theorem which applies in the case of passive, but not active, networks.

However, other representations are possible. All that is required of a representation is that we should be able to form two equations of the form shown above in which all four $h$-parameters are different. One such alternative representation containing two passive and two active elements is shown in Fig. 13. The constants of this circuit can, of course, readily be worked out in terms of $h$ or any other set of parameters. In particular, the factor " $b$ " for the grounded-base configuration is very nearly equal to 1 .
" $a$ " and " $a$ dashed ".-As was mentioned at the start, $a$ and $a^{\prime}$ are two additional parameters loosely termed "current gain" in grounded-base and grounded-emitter configurations, respectively.

Current gain under what conditions?
With the output short-circuited to a.c., i.e., with $\mathrm{dV}_{2}=0, a=\left(\mathrm{dI}_{e} / \mathrm{dI}_{e}\right)$ in the grounded-base configuration

$$
=-h_{21}=\left(r_{b}+r_{m}\right) /\left(r_{b}+r_{c}\right)
$$

while $a^{\prime}=\left(\mathrm{dI}_{c} / \mathrm{dI}_{b}\right)$ in the grounded-emitter configuration

$$
=h_{21}^{\prime}=\left(r_{m}-r_{e}\right) /\left(r_{e}+r_{e}-r_{m}\right)
$$

and there is no relation whatever between them.
On the other hand, if the input is short-circuited



Fig. 14. Definition of the Mullard transistor parameters.
to a.c., i.e., $\mathrm{dV}_{1}=0$ and the currents are varied by varying the output voltage, $a=\left(\mathrm{dI}_{c} / \mathrm{dI}_{e}\right)$ in the grounded-base configuration

$$
=\left(h_{11} h_{22}-h_{12} h_{21}\right) / h_{12}=\left(r_{b}+r_{e}\right) / r_{b} .
$$

and $a^{\prime}=\left(\mathrm{dI}_{c} / \mathrm{dI}_{b}\right)$ in the grounded-emitter configuration

$$
\begin{aligned}
& =-\left(h_{11}^{\prime} h_{22}^{\prime}-h_{12}^{\prime} h_{21}^{\prime}\right) / h_{12}^{\prime}=-\left(r_{b}+r_{e}\right) / r_{B} \\
& =a /(1-a) .
\end{aligned}
$$

We note that $a$ in this case is greater than 1 , and so $a^{\prime}$ is negative. This latter case is not equivalent to using the transistor the reverse way round.

When treating a transistor as a T-network, we find $a$ quoted as $r_{m} / r_{c}$ and $a^{\prime}$ given as $r_{m} /\left(r_{c}-r_{m}\right)=$ $a /(1-a)$. In fact, in order that the current gain may be $r_{m} / r_{c}$ in the grounded-base configuration, it is necessary to measure it with a collector load of $r_{b}\left(1-r_{c} / r_{m}\right)$ in circuit (about $-17 \Omega$ !), whilst to get a current gain of $r_{m} /\left(r_{c}-r_{m}\right)$ in the groundedemitter configuration, you would need a collector load of $\left(r_{c} r_{e} / r_{m}\right)$, that is, about $18 \Omega$.

In view of the constants of a normal transistor, this ambiguity is perhaps rather academic. Taking $r_{b}$ as $700 \Omega, r_{c}$ as $1 \mathrm{M} \Omega, r_{e}$ as $18 \Omega$ and $r_{m}$ as $976 \mathrm{k} \Omega$, the short-circuit current gain in the groundedbase configuration $=\left(r_{b}+r_{m}\right) /\left(r_{b}+r_{e}\right)=976.7 /$ 1000.7, whilst the usual expression, $r_{m} / r_{c}$, for current gain in this configuration comes to 0.976 .

Likewise, the short-circuit current gain in the grounded emitter configuration $=\left(r_{m}-r_{e}\right) /\left(r_{e}+\right.$ $\left.r_{c}-r_{m}\right)=975.982 / 24.018$, whilst the usual expression, $r_{m} /\left(r_{c}-r_{m}\right)$, for current gain in this configuration comes to 976/24.

However, it is as well to be on one's guard when discussing current gain.

Mullard Parameters.-Mullard parameters are the input and output resistances of a transistor with output and input respectively open-circuited and short-circuited to a.c.

The symbols for the grounded-base parameters are $r_{i,}, r_{0}, r_{11}$ and $r_{22}$, Corresponding dashed and double-dashed symbols are defined for the other two configurations. The parameters are defined in Fig. 14.

Mullard parameters perhaps betray the influence of the electrical engineer, who tends to define the properties of all his machinery in terms of its open and short circuit impedances or their equivalent, but they have the great advantage of being directly measurable. A large inductance is used to provide an open circuit to a.c. if a d.c. path is required, and a large condenser is used to provide a short circuit to a.c. where a d.c. open circuit is required.

Mullard parameters, however, suffer from one vital limitation. Unlike the other two sets of parameters, they fail in themselves to give all the properties of the transistor.

The reason is very simple. Consider for example the grounded-base configuration:-
$r_{i}=$ input resistance with output short-circuit to a.c., i.e. with $\mathrm{dV}_{2}=0$,

$$
=h_{11}
$$

$r_{o}=$ output resistance with input short-circuit to a.c., i.e. with $\mathrm{dV}_{1}=0$,
$=h_{11} /\left(h_{11} h_{22}-h_{12} h_{21}\right)$
$r_{11}=$ input resistance with output open-circuit to a.c., i.e. with $d \mathrm{I}_{2}=0$,

$$
=\left(h_{11} h_{22}-h_{12} h_{21}\right) / h_{22} \text { and }
$$

$r_{22}=$ output resistance with input open-circuit to a.c., i.e. with $\mathrm{dI}_{1}=0$,

$$
=1 / h_{22}
$$

It is seen at once that $r_{i} r_{22}=r_{0} r_{11}$.
In the same way, similar relations may be obtained for the other two configurations.

Thus, although you can measure four constants of the transistor, you only obtain three separate


Fig. 15. Equivalent circuits of a transistor using Mullard and $T$-network parameters.
pieces of information about it; it is therefore necessary to measure one other property of the transistor, for instance the short-circuit current gain in the grounded-emitter configuration, $h^{\prime}{ }_{11}$.

Be that as it may, Mullard parameters are defined, and can be very easily related to the other sets of parameters. We limit ourselves to a derivation of only one of these relations to show the method:$r^{\prime \prime}{ }_{0}$ in terms of the T -network parameters.

Fig. 15 gives the equivalent circuits, from which we have:-
$V_{e b}=r_{e} \mathrm{I}_{e}+r_{b} \mathrm{I}_{b}=r_{e} \mathrm{I}_{e}+r_{e}\left(\mathrm{I}_{e}-\mathrm{I}_{b}\right)-$ $r_{m} \mathrm{I}_{e}$
Therefore

$$
\mathrm{I}_{b}=\mathrm{I}_{e}\left(r_{c}-r_{m}\right) /\left(r_{b}+r_{c}\right)
$$

Therefore

$$
\mathrm{V}_{e n}=\mathrm{I}_{e}\left(r_{e}+r_{b}\left(r_{c}-r_{m}\right) /\left(r_{b}+r_{c}\right)\right)
$$

and so

$$
\mathrm{V}_{e b} / \mathbf{I}_{e}=r_{i}=r_{o}{ }^{\prime \prime}=r_{e}+r_{b}\left(r_{c}-r_{m}\right) /\left(r_{b}+r_{e}\right)
$$

We give a table of various parameters in terms of $h^{\prime}{ }_{21}$ and the Mullard parameters for the groundedemitter configuration, since the latter are those most easily measured:-
T-network Parameters
$r_{b}=\left[\left(1+h_{21}^{\prime}\right) r_{11}^{\prime}-r_{i}^{\prime}\right] h_{21}^{\prime}, r_{c}=\left(1+h_{21}^{\prime}\right)$
$r_{22}^{\prime}, r_{e}=\left(r_{i}^{\prime}-r_{11}^{\prime}\right) / h_{21}^{\prime}$ and $r_{m}=h_{21}^{\prime} r_{22}^{\prime}-$
$\left(r_{i}^{\prime}-r_{11}^{\prime}\right) / h^{\prime}{ }_{21}$
Grounded-Base Mullard Parameters
$1 / r_{i}=\left(1+h_{21}^{\prime}\right) / h^{\prime}{ }_{21} r_{o}^{\prime}+\left(1+h^{\prime}{ }_{11}\right) / r_{i}^{\prime}-$
$1 / h^{\prime}{ }_{21} r_{22}^{\prime}, r_{0}=r^{\prime}{ }_{0}, r_{11}=r_{11}^{\prime}$ and $r_{22}=r^{\prime}{ }_{0} r^{\prime}{ }_{11}$
(1/ris) (q.v.)
Grounded-Collector Mullard Parameters
$\begin{aligned} & r_{i}^{\prime \prime}=r_{i}, r_{0}^{\prime \prime}=r_{i} \text { (q.v.), } r_{11}^{\prime \prime}=r_{22} \text { (q.v.) and } \\ & r^{\prime \prime}{ }_{22}=r_{22}^{\prime}\end{aligned}$
$h_{11}=r^{\prime}{ }_{i}\left(r^{\prime}{ }_{22} / r_{22}\right)$ (q.v.), $h_{12}=1-\left(1+h^{\prime}{ }_{21}\right)$
$\left(r^{\prime}{ }_{22}^{\prime} r_{22}\right)(\mathrm{q} . \mathrm{v}),. h_{21}=-\left(h_{21}^{\prime}+\left(r_{i}^{\prime} / r_{0}^{\prime}\right)\right)\left(r^{\prime}{ }_{22} / r_{22}\right)$
(q.v.) and $h_{22}=1 / r_{22}$ (q.v.)
$h_{11}^{\prime}=r_{i,}^{\prime}, h_{12}^{\prime}=\left(r_{i}^{\prime}-r_{11}^{\prime}\right) / h^{\prime}{ }_{21} r^{\prime}{ }_{22}$ and $h_{22}^{\prime}=$
$1 / r^{\prime}{ }_{22}$
Grounded-Collector h-parameters
$h^{\prime \prime}{ }_{11}=r^{\prime}{ }_{i}, h^{\prime \prime}{ }_{12}=1-h_{12}^{\prime}$ (q.v.), $h^{\prime \prime}{ }_{21}=$
$-\left(1+h_{21}^{\prime}\right)$ and $h^{\prime \prime}{ }_{22}=1 / r^{\prime}{ }_{22}$

Conclusion.-The whole of this article has been written around the transistor and a number of approximations may therefore be made by reason of its nature. The treatment, however, has been quite general and is therefore equally applicable to any quasi-linear active four-terminal network.

## SHORT-WAVE CONDITIONS Prediction for June



G.M.T.

G.M.T.

THE full-line curves indicate the highest frequencies likely to be usable at any time of the day or night for :eliable communications over four long-distance paths from this country during June.

Broken-line curves give the highest frequencies that will sustain a partial service throughout the same period.
........... FREQUENCY BELOW WHICH COMMUNICATION SHOULD BE POSSIBLE
FCR $25 \%$ OF THE TOTAL TIME

-     - PREDICTED MEDIAN STANDARD MAXIMUM USABLE FREQUENCY
— FREQUENCY BELOW WHICH COMMUNICATION SHOULO BE POSSIBLE ON ALL UNDISTURBED DAYS


## Stereo Broadcasting

ON April 20th the U.S. Federal Communications Commission authorized the introduction of f.m. multiplex stereo broadcasting from June 1st. The technical standards approved are a combination of proposals put forward by the Zenith Radio Corp. and General Electric. Other systems investigated by the American Electronic Industries Association at the request of the F.C.C. included those of Crosby Teletronics Corp., Multiplex Development Corp., Calbest Electronics Co. and also the E.M.I. Percival system (see W.W. November, 1958).

The approved system, which is fully compatible with single-channel f.m. broadcasting, is f.m./a.m. with the sub-carrier amplitude modulated.

## New F.M. Stations

WORK is well advanced on the first two of the ten additional v.h.f. sound broadcasting stations announced by the B.B.C. in 1959. They both share sites with existing television stations-at Londonderry, Northern Ireland, and Les Platons, Jerseyand their directional aerials will be erected on the stations' television masts.

Both stations are planned to be brought into operation during this summer. The Londonderry station, which will have a mean e.r.p. of 5.6 kW , will radiate on $88.3 \mathrm{Mc} / \mathrm{s}$ (Light), $90.5 \mathrm{Mc} / \mathrm{s}$ (Third), and $92.7 \mathrm{Mc} / \mathrm{s}$ (Home). Les Platons, which will have a mean e.r.p. of 3.1 kW , will transmit on $91.1 \mathrm{Mc} / \mathrm{s}$ (Light), $94.45 \mathrm{Mc} / \mathrm{s}$ (Third), and 97.1 $\mathrm{Mc} / \mathrm{s}$ (Home). At Les Platons the programmes will be received by radio from either the North Hessary Tor or Rowridge v.h.f. stations.

## Technical Writing Scheme

MODIFICATIONS have been made to the Technical Writing Award Scheme which is now jointly sponsored by the Radio Industry Council, the Electronic Engineering Association and the Electronic Valve and Semi-Conductor Manufacturers' Association. In future the number of annual awards will be increased from six to eight and their value increased from 25 to 30 guineas each. Also some of them will be awarded for purely technical articles in addition to those made for articles of " commercial or interpretative merit."

The judges for 1961 will be H. E. F. Taylor (E.E.A.), who is chairman, B. C. Brookes (University College, London), A. H. Cooper (E.M.I.), F. Jeffery (Murphy), Professor C. W. Oatley (Cambridge University), G. Reeves (A.E.I.), and Dr. R. C. G. Williams (Philips).

## Band V Colour

COLOUR transmissions on 625 lines are to be radiated experimentally in Band $V$ by the B.B.C. from the Crystal Palace station. E.M.I. Electronics
are carrying out the necessary modifications to the $10-\mathrm{kW}$ transmitter which was used for the 405- and 625 -line monochrome test transmissions in Band V during 1957 and 1958. No date has yet been announced for these transmissions nor has it been stated whether the sense of vision modulation will be positive or negative or if a.m. or f.m. will be used for sound. It is understood that an $8 \mathrm{Mc} / \mathrm{s}$ bandwidth will be employed.

The P.M.G., in reply to a question in the House on May 10th, reiterated his decision not to accede to the request of the B.B.C. for a limited service of colour television on 405 lines.

## U.H.F. Television in Germany

WEST GERMANY'S first regular u.h.f. television service was inaugurated by Hesse Radio, Frankfurt, on May 1st. Other West German broadcasting organizations are planning to start u.h.f. services in June. Hesse Radio is using some of the 30 u.h.f. stations sct up by the Government in preparation for the start of the "Adenauer" network, which, it will be remembered, was finally banned by a decision of the Federal Court earlier in the year.

A second programme is being introduced by each of the broadcasting organizations on u.h.f., but it is planned to combine their operations from June 1962 into a network with a proportion of local progranmes.

Italy's second TV service is scheduled to come into operation at the beginning of November. It will be operated by the national broadcasting organizationRadiotelevisione Italiana (RAI)-whose present TV network totals nearly 400 stations, the majority of which are low-power relay transmitters. About $20 \%$ of last year's income of RAI was from advertising which is carried on by both the sound and television services.

Dual TV Standards.-France, which has used the S19-line standard for its television service since 1950, is introducing the 625 -line standard for its second service scheduled to start in June next year in the Paris region. Stations will operate in Band IV (470$582 \mathrm{Mc} / \mathrm{s}$ ).

Plastics Exhibition.-Over 350 U.K. and overseas manufacturers of plastics materials, finished products and machincry will be exhibiting at the biennial International Plastics Exhibition which opens at Olympia, London, on June 21 st for 10 days. A three-day convention (June 26th-28th) is being held in conjunction with the Exhibition and admission to both is by ticket obtainable free from British Plastics (Dorset House, Stamford Street, London, S.E.1), which, with the cooperation of the British Plastics Federation, is organizing the show.

Berlin Radio Show.-The German national radio and television show is to be held for the first time for 22 years in Berlin from August 25th to September 3rd. There will be 150 exhibitors in the 13 halls in the exhibition grounds at the Funkturm in West Berlin. The Federal Post Office is issuing a special 20-pfennig stamp to mark the occasion.
B.R.E.M.A. Council.-The new council of the British Radio Equipment Manufacturers' Association consists of the following firms, whose representatives' names are in parentheses:-A. J. Balcombe (E. K. Balcombe), British Radio Corporation (F. W. Perks), Bush Radio (G. Darnley-Smith), Ferguson Radio Corporation (S. T. Holmes), General Electric Co. (M. M. Macqueen), Kolster-Brandes (E. P. Wethey), Murphy Radio (A. P. Power), Perdio (Viscount Suirdale), Philips Electrical (A. L. Sutherland), Radio and Allied Industries (R. H. Pengelly), Rediffusion Vision Service (M. Exwood), and Ultra Radio and Television (E. E. Rosen). A. L. Sutherland is again chairman of the association.

British Wireless Dinner Club.-At the 38th annual dinner of the Club on April 14th Air Marshal Sir Leslie Dalton-Morris, was elected president in succession to Admiral Sir Alan Scott-Moncrieff. Sir Leslie was Air Officer C.-in-C., Signals Command R.A.F. until his appointment earlier this year to the Maintenance Command. The new vice-president of the Club is MajorGeneral L. de M. Thuillier, who headed the U.K. team which towards the end of last year visited the U.S.A. for discussions on the use of earth satellites for communications systems. The joint honorary secretaries of the Club are Capt. F. J. Wylie, director of the Marine Radio Advisory Service, and L. T. Hinton, of Standard Telephones and Cables.
B.E.A.M.A.-An Industrial Electronic Equipment Section has been formed by the British Electrical and Allied Manufacturers' Association. In addition to the day-to-day work relating to technical matters, standardization, exports, and statistics, the scction is intended also to " provide a means of closer discussions with other associations and help towards the wider examination of general policy questions affecting the industrial electronics industry as a whole." W. Gregson, of Ferranti, has been elected chairman, and R. J. F. Howard, of Metal Industries, is vice-chairman. A. Newton has been appointed secretary.

Society of Relay Engineers.-R. I. Kinross, managing director of Redifiusion Research, Ltd., has been reelected president of the Society of Relay Engineers, and R. P. Gabriel, chief engineer of Rediffusion, Ltd., reelected vice-president. The secretary of the society is T. H. Hall, Obelisk House, Finedon, Northants.


ELEGANCE. - The Ultra TR70 transistor portable which was selected by the Council of Industrial Design for the Duke of Edinburgh's Award for Elegant Design. It employs seven transistors and two crystal diades (one of which provides additional a.g.c.) and covers both m.w. and l.w. bands. The dimensions are $12 \times 8 \frac{1}{4} \times 3 \mathrm{in}$.

Television Society Council--At the annual general meeting of the Television Society on April 27th the following were elected to fill the vacancies on the council:-B. Eastwood (A.E.I.), I. J. P. James (E.M.I.), C. A. Marshall (British Communications and Electronics) and Dr. R. D. A. Maurice (B.B.C.). Three new vicepresidents were elected. They are Sir Willis Jackson, Dr. R. L. Smith-Rose and P. H. Spagnoletti.
R.E.C.M.F.-John Clark (Plessey Company) has been elected vice-chairman of the Radio and Electronic Component Manufacturers' Federation, in succession to Dr. G. A. V. Sowter (Telcon Metals) who, as announced last month, is now chairman.

Stockholm R.E.C.M.F. Show.-The date of the fourth British Components Show to be held in Stockholm should be added to the list of exhibitions on page 337. Organized by the Radio and Electronic Component Manufacturers' Federation it will be held from October 9 th to 13 th.

Milan.-Another addition should be made to the list of exhibitions on page 337. It is the date of the Milan Radio Show which will be held from September 10th to 17 th.

Baird Memorial Lecture.-Professor Martin Ryle, Professor of Radio-Astronomy at Cambridge University, is to deliver the second Baird Memorial Lecture at the Royal College of Science and Technology, Glasgow, on June 7th. His lecture is entitled "Exploring the Universe with Radio Waves."
Industrial Electronics.-C. Metcalfe, of E.M.I., who is chairman of the Electronic Forum for Industry, will be one of the speakers at the Industrial Electronics Symposium in Boston on September 20th and 21st.

## CLUB NEWS

Barnet.-At the May 30th meeting of the Barnet and District Radio Club V. A. Frisbee (G3KVF) will talk about mobile equipment. On June 27th T. H. A. Withers (G3HGE) will deal with v.h.f. equipment. Mectings are held at 8.0 at the Red Lion Hotel, Barnet.

Bexleyheath.-The North Kent Radio Society will be operating the club transmitter (G3ENT/A) at the Careers Exhibition being held in the Crayford Town Hall in connection with the Commonwealth Technical Training Week (May 29th-June 3rd). It will be using telephony on all bands but mainly on 1.8 and $7 \mathrm{Mc} / \mathrm{s}$.
Birmingham.-"Transistors" is the title of the talk to be given by N. B. Simmonds to the Slade Radio Society on June 16th at 7.45 at Church House, High Street, Erdington.
East London.-Short-wave listeners in the London districts of East Ham, Manor Park, IIford and Barking are invited to join the East Ham Group of the Radio Society of Great Britain. Details of activities are obtainable from J. Brown (G3DBO), 19, Compton Avenue, London, E.6.
Halifax.-The Halifax and District Amateur Radio Society now meets at the Beehive and Crosskeys Inn at 7.45 on alternate Tuesdays. The June meetings are on the 6th and 20th. The town's second radio club has recently been formed and is known as the Northern Heights Amateur Radio Society. It meets on alternate Wednesdays at 7.45 at the Sportsman Inn, Ogden.
Harlow- - A mobile rally is being organized by the Harlow and District Radio Society for June 11 th. It will be held at Magdalen Laver, near Harlow, Essex. There will be two "talk-in" stations-G3ERN on 160 m and G3JMA on 2 m .
Sheffield Amateur Radio Club meets on the 2 nd and 4 th Wednesday of cach month at 7.45 at the Dog and Partridge, Trippett Lane. Details of forthcoming meetings are available from D. R. A. Hill, 16 Tylney Road, Sheffield, 2.
Warminster.-On June 25th the City and County of Bristol Group of the R.S.G.B. is holding a mobile rally at Longleat House, near Warminster, Wiltshire. The control stations will operate on $1.9 \mathrm{Mc} / \mathrm{s}$ ( $\mathrm{G} 3 \mathrm{CHW} / \mathrm{A}$ ) and 144.15 $\mathrm{Mc} / \mathrm{s}(\mathrm{G} 3 \mathrm{GYQ} / \mathrm{A})$ ), Particulars from C. N. Chapman (G2HDR), "Yeovil," Stoke Hill, Stoke Bishop, Bristol, 9.

## Personalities

Sir Harold Roxbee Cox, D.Sc., Ph.D., D.I.C., who is 58, has been appointed chairman of the Council for Scientific and Industrial Research for five years from October 1st. He will succeed Sir Harry Jephcott, who has been chairman of the Council since its formation in 1956. Sir Harold has been a member of the Council since 1957. He spent over twenty years in the Government service prior to 1954 when he became associated with a number of industrial companies. He is chairman of the National Council for Technological Awards, and vice-chairman of the Governors of the College of Aeronautics.
W. J. Richards, C.B., C.B.E., director of the Royal Radar Establishment, Malvern, since 1953, has been appointed director of the proposed new Staff College for Further Education. He will take up his new appointment in July. The aim of the College, premises for which have yet to be found, is to create a "national centre at which principals and other selected staff from colleges of technology and commerce can exchange ideas and experience with senior staff from industry, local and central government, the universities and overseas and where they can jointly study the aims, ideas and the growth of education in the light of the rapid advances in science and technology." Mr. Richards, who is 57, graduated in engineering with first-class honours from Manchester University. After spending some years in research and development work at the R.A.E., Farnborough, he became deputy director of scientific research at the Ministry of Aircraft Production in 1942. During his tenure of office at R.R.E. he has established the College of Electronics.

W. J. Richards


Dr. R. A. Bones

Roger A. Bones, B.Sc., Ph.D., A.Inst.P., has joined Wayne Kerr Laboratories as head of the company's contracts division. After graduating at London University in 1949 with first-class honours in physics and mathematics, Dr. Bones studied X-ray crystallography at University College, London. After a period with the Foreign Office, where he was concerned with electronics and telecommunications research, he lectured at Hong Kong University for three years, later becoming an industrial adviser to the Federal Ministry of Commerce and Industries in Nigeria. Immediately prior to joining Wayne Kerr, Dr. Bones was with the de Havilland Aircraft Company.
D. J. Cole, B.A., LL.B., son of the founder and present chairman of E. K. Cole, Ltd., has been appointed to the board of directors. He joined the company in 1954.
R. Wrathall, formerly radio and television sales manager of E. K. Cole Ltd., has been appointed director and general manager of Ekco Radio and Television Ltd, which has been formed to control the marketing of all Ekco television and radio receivers, car radio and allied

R. Wrathall

R. A. Drummond
products. W. M. York is chairman of the board of directors. Three additional directors have been appointed to the board of the associated company Ferranti Radio \& Television Ltd. R. A. Drummond, who has been with the Ekco Group for nearly 30 years and has been sales manager of Ferranti Radio and Television Ltd. since its formation four years ago, has become director and general manager. The other new directors are N. T. Atkinson and A. C. Segger, who have also joined the board of Ekco Radio \& Television.
J. W. Soulsby has been re-elected chairman of the Radio Officers' Union for his seventh consecutive term of office. Born in 1900, he joined Marconi's in 1918, and is at present Chief Radio Officer in the British India Steam Navigation Company's Uganda. The new vicechairman of the Union is J. G. Salveson, who is a radio officer with British European Airways, which he joined soon after demobilization from the R.A.F. in 1946. He is 41 .
K. J. H. Adams, recently appointed production manager of the semiconductor division of Brush Crystal Company, Hythe, Southampton, was formerly senior production engineer at Semiconductors Lid. After graduating with honours in physics at London University in 1951, he joined Mullard Radio Valve Company as a development engineer working on problems associated with repeater valves. In 1953 he left the company but returned the following year to head the department handling its guided missile valve project. He joined Semiconductors in 1959.
J. W. C. Robinson, M.B.E., managing director of Relay Exchanges Ltd., has been appointed chairman of the company in succession to the late Sir Walter J. Womersley, Bart. Mr. Robinson is also joint managing director of Goodmans Industries, which was acquired by Relay Exchanges in 1958.
E. G. Westray has retired after 49 years' service with Ericsson where, since 1946, he has been export sales manager. Mr. Westray, who was for many years in the company's laboratories, was throughout the last war Ericsson's liaison officer to the Ministries.
F. A. Vick, O.B.E., Ph.D., who was appointed director of the Atomic Energy Research Establishment at Harwell last September, has also been appointed director of the Research Group of the U.K. Atomic Energy Authority. Dr. Vick, who is 49, was Professor of Physics in the University of North Staffordshire from 1950 until 1959 when he joined the A.E.R.E. as deputy director. Prior to that he was for five years on the staff of Manchester University and throughout the war was Assistant Director of Scientific Research in the Ministry of Supply.

Air Commodore John C. Millar, D.S.O., the new commandant of the R.A.F. Central Signals Establishment, was formerly Director of Flight Safety at the Air Ministry. Air Commodore Millar, who is 47 and was educated at Trinity College, Cambridge, was Chief Signals Officer at R.A.F. Bomber Command in 1954.

Colin H. Gardner, manager of the Mullard Films and Lectures Organization for the past eight years, has retired after thirty-two years' service with the company. He will continue to act in an advisory capacity to the organization. He has been succeeded by P. I. Nicholson, M.Brit.I.R.E., who joined the company in 1950 and for the past two years has been deputy manager of the Films and Lectures Organization.

Arthur L. Budlong (WIBUD), secretary and general manager of the American Radio Relay League, and editor of QST, has retired and is succeeded by John Huntoon (WILVQ). "Bud," as he is affectionately known in amateur circles, is also secretary to the International Amateur Radio Union. He has been on the staff of the A.R.R.L. for 37 years.

Peter Rushen, who writes in this issue on leakage in printed circuits, has been with Bush Radio since 1946, except for a two-year break when he was with English Electric. Whilst at English Electric he worked on the Thunderbird ground-to-air missile. He is at present engaged in the development of export television receivers.

Graham Miller, B.Sc., has been appointed home sales manager of Wayne Kerr Laboratories, which he joined two years ago. He studied electronics at Manchester College of Technology and was a graduate in physics at Swansea University. He subsequently became a circuit design engineer and spent two years as head of the Ferranti Standards Laboratory at Wythenshawe.

G. Miller

H. G. Hinckley
H. G. Hinckley has been appointed manager of the Machine Tool Control Department of Ferranti, Ltd., Edinburgh, in place of D. T. N. Williamson, manager of the department since its inception in 1952 . Mr. Hinckley, who is 40 , served with T.R.E. and British European Airways before joining Ferranti in 1948. Mr. Williamson, who has joined Molins Machine Co., as director of research and development, will be remembered as the designer of the amplifier bearing his name which he described in Wireless World in 1947

Brian M. Lee, who a few months ago was appointed manager of the Industrial Division of Belling \& Lee Ltd., has become a member of the board of executive directors.

# News from Industry 

British Space Development Company, which, as announced in our March issue, has been formed by a consortium of companies, has appointed Sir Robert Renwick as chairman and the following have been elected to the board: J. R. Brinkley (Pye), Grp. Capt. E. Fennessy (Decca Radar), G. C. I. Gardiner (Hawker Siddeley Aviation), A. A. Rubbra (Rolls-Royce) and Grp. Capt. D. Saward (Rank Organization). The following companies, together with those already mentioned, are founder members: A.E.I., Associated Television, B.I. Callender's Cables and Plessey. A technical committee has been set up "to examine and prepare plans designcd to achieve British participation" in the future utilization of "space." The members of the Technical Committee, of which G. K. C. Pardoe is chairman, are: D. E. Burchett (A.E.I.), A. V. Cleaver (Rolls-Royce), J. M. C. Dukes (Plessey), Dr. W. F. Hilton (Hawker Siddeley), W. M. Lloyd (Rank), L. F. Mathews (A.T.V.), Dr. K. Milne (Decca Radar), T. P. Blott (Pye) and G. A. Dodd (B.I.C.C.).

Vactric.-The News of the World Organisation Ltd., has purchased the whole of the issued share capital of Vactric (Control Equipment) Ltd., and Vactric (Precision Tools') Ltd., from the receiver of Vactric Ltd (in liquidation). New boards will be formed with Sir William Carr as chairman, and the present managing director, T. W. Roberts, continuing in that capacity.
G.E.C.-Sobell-McMichael.-The merger between the General Electric Company and Radio \& Allied (Holdings) has now been completed and M. Sobell, chairman of R. \& A., and A. Weinstock, the managing director, have joined the board of G.E.C. Lord Coleraine has also joined the board and Sir Toby Low has been elected vice-chairman. The object of the merger is to "strengthen the operations of the two companies in the radio and television field and in domestic appliances." Each company will continue to market its own range of receivers and to this end a new G.E.C. company, G.E.C. (Radio \& Television) Ltd., has been formed.

The headquarters are at Langley Park, Slough (Tel.: Slough 22201). The service department is at Lena Gardens, London, W. 6 (Tel.: Riverside 4671).

Relay Exchanges Ltd. announce a group trading profit for 1960 of $£ 3,949,892$ which is almost $£ 600,000$ above the 1959 figure. After deducting £2,747,397 for depreciation and further sums for taxation, the net group profit was $£ 1,024,825$.

Elliott-Automation.-Preliminary figures for 1960 issued by Elliott-Automation Ltd., show a group profit before taxation of $£ 1,776,510$ compared with $£ 1,015,630$ the year before. The recently acquired Rheostat Company Group contributed $£ 520,648$ towards the 1960 total.

Ultra have entered into two agreements with companies in the western hemisphere. Ultra Electronics has signed an agreement with Electronics Investment Management Corp., of the U.S.A., for the interchange of information on research, development, marketing and production of electronic equipment. The corporation is associated with a number of companies both in the U.S.A. and outside to which it renders advisory and consulting services. The parent company Ultra Electric (Holdings) has entered into an agreement with Electronics International Capital Ltd., of Bermuda, whereby E.I.C. has purchased $40 \%$ of the shares of Ultra Electronics Ltd. for a sum of $£ 800,000$. Electronics International Capital Ltd. has, in addition, agreed to make available five-year loan facilities up to a total of £950,000.
B.S.R. announce record production and sales of all their products in 1960 resulting in a group net profit of $£ 860,452$ after deducting $£ 666,124$ for tax. The net profit was some $£ 73,000$ up on 1959 . The group recently purchased for cash majority shareholdings in two companies, Tape Heads Ltd. (formerly Bradmatic Productions Ltd.), of Birmingham, and P. A. Marriott \& Company Ltd., of Wembley, Middlesex. The managing directors of these two companies-G. Littlewood and P. A. Marriott, respectively-will continue in office.

Southern Areas Electric Corp., of which Ross, Courtney \& Co. is a subsidiary, record a profit for 1960 of $£ 47,782$ (after taxation) which was $£ 4,000$ above the 1959 figure.

Plessey-Mallory.-Under an agreement with P. R. Mallory and Company Inc., the Plessey Company is now manufacturing a range of solid electrolyte tantalum capacitors. Chief features of the new capacitors are stability of capacitance with changes of temperature, and the ability to operate at extremely low temperatures down to $-80^{\circ} \mathrm{C}$ without loss of performance. The range of capacitance values available is 0.33 to 330 microfarad at d.c. working voltages of 35 to 6 .

Bach-Simpson Company, instrument manufacturers of London, Ontario, have appointed Aveley Electric Ltd., of South Ockendon, Essex, as their sole U.K. agents.

Marconi's W.T. Company has acquired the Wembley factory of E.M.I. Electronics Ltd. as a staffed and equipped working entity. The factory is approximately $85,000 \mathrm{sq} \mathrm{ft}$ and at present employs between $400-500$ people.

Cossor Radar \& Electronics.-As a result of severe damage by fire to its Servicing Department at West Norwood, Cossor Radar and Electronics have transferred the department to Edinburgh Place, Temple Fields, Harlow, Essex (Tel.: Harlow 25537).

Western Electric Company, manufacturing and supply unit of the Bell Telephone System, of the United States, has charged Transitron Electronic Corp., of Wakefield, Mass., with infringement of five separate Bell System U.S. Patents relating to semiconductor devices such as transistors and silicon diodes.
Aveley.-The recently formed Components Division of Aveley Electric Ltd. is holding an exhibition of its products in the Grill Foyer of the Kensington Palace Hotel, London, W.8, from May 30th to June 2nd.

Perth Padios Manufacturing Ltd., which was formed in 1953 by Mohamed Hussain Ismail, has gone into liquidation. Since 1957 the company has been operating from Marten House, 39-47 East Road, London, N.1. The liquidator is A. J. Cooke, 9 Basinghall Street, London, E.C. 2.
CQ Audio, formerly R.G.A. Sound Services, of 111 Cecil Road, Enfield, Middx., have gone into liquidation.
C.R.T. Servicing--Facilities for the servicing of Ediswan Mazda cathode-ray tubes, formerly available at A.E.I. Radio and Electronic Components Division's service department at Brimsdown, Middx., have been moved to the division's factory at Ducklees Lane, Ponders End, Middx. Bulbs for salvage under A.E.I.'s reclamation scheme should be returned to the same address.

## EXPORTS

B.E.A.M.A. will hold a two-day export conference on October 5th and 6th. The opening speaker will be the Rt. Hon. Reginald Maudling, President of the Board of Trade.

Marconi Instruments' mobile demonstration unit is now on a 3,000 -mile tour of Germany and Scandinavia. The vehicle houses over 30 electronic measuring instruments.

New Zealand.-Le Quesne Electronics Ltd., of 3, Faraday Street, Napier, New Zealand, wish to represent British radio and electronic equipment and component manufacturers.

Swedish Agents.-Telequipment have appointed their Swedish agents Magnetic AB, Stora Ny Gatan, Stockholm 3.

New Zealand Agents.-Telemechanics Ltd., of Southampton, have appointed Electronic Development and Applications Ltd., P.O. Box 1905, Auckland, C.1, as their agents in New Zcaland.

Portuguese Agents.-Telemechanics have appointed Rualdo Lda., Rua S. Jose 15-1, Lisbon 2, Portugal, as their agents.

## BRITISH TRADE FAIR moscow

THE radio and electronics industry is well represented among the 600 or more companies participating in the first British Trade Fair to be held in Moscow which opened on May 19th for a fortnight. The following companies and research establishments are among the exhibitors in the fourteen sections, one of which is devoted to electrical equipment (which includes radio and television consumer and capital goods and electronic equipment) and another to scientific instruments.

This fair will be followed in July by a reciprocal Soviet Trade and Industrial Exhibition to be held at Earls Court, London. Both exhibitions are sponsored jointly by the Association of British Chambers of Commerce and the All-Union Chamber of Commerce of the U.S.S.R.

Advance Components
Associated Electrical Industries Automatic Telephone \& Electric Co. Avo
Blackburn Electronics
Board of Trade
British Ferrograph Recorder Co.
Brush Electrical Engineering Co.
Cambridge Instrument Co .
Casella (Electronics)
Cossor Instruments Crompton Parkinson Dawe Instruments E.M.I. Electronics

Electronic Instruments Elliott Automation Group Endecotts (Filters) Endecotts (Filte
English Electric
English Electric Valve Co.
English
Faraday
Faraday
Fleming Radio (Developments) G. B. Kalee

Goodmans Industries
Griffin \& George

Kelvin \& Hughes Labgear
Lancashire Dynamo Holdings
Langham Thompson, J.
Marconi Instruments
Marconi International Marine
Marconi's Wireless Telegraph Co
Mullard Equipment
Napier, D., \& Son
Pye
Pye T.V.T.
Pye Telecommunications
Pye, W. G., \& Co.
Rank Cintel
Savage, Bryan
Scientific Instrument Manufacturers Assoc.
Solartron Electronic Group
Standard Telephones \& Cables
Taylor, Taylor \& Hobson
Telequipment
Thermionic Products (Electronics) Ulera Electronics
Wayne Kerr Laboratories

# Reliable Circuit Designs 

## REDUCTION OF DEPENDENCE ON VALVE CHARACTERISTICS

By B. PRIESTLEY

THE object of this article is to suggest some general pointers to good design procedures which can be used to produce a required standard of performance in most circuits. Electronic circuit design is not a black art, giving unpredictable results, but a logical science capable of producing reliable and stable performance. By reliability is meant stability in the face of variations both in circuit components, due to tolerances, ageing and temperature, and in external supplies.

The extent to which good design can help reliability does not always seem to be realized. This may be due to many textbooks giving explanations of how a particular circuit works, once constructed, but no explicit design information.

Circuits which depend on ill-defined parameters should be avoided. Thus, for example, one can design a blocking oscillator for a specific pulse width, but with no great accuracy. If an accurate pulse width is required it is preferable to define it with a delay line.

The effects of supply voltage variations and component tolerances are calculable and should be evaluated in all cases where high reliability is required. This, of course, does not reduce the effects of variations, but calls attention to the variation in performance to be expected and may spotlight defects such as working with unequal safety margins for under- and over-voltage, or decide between two apparently equally suitable circuits.

Some components, particularly valves and transistors, have parameters which are subiect to wide tolerances, made wider by variations due to age and temperature respectively. While more information on the form of variation to be expected would be welcome, much can be done by the application of negative feed-back and similar techniques to make the operating conditions largely independent of valve and transistor parameters.

In order to illustrate the above points, two similar cathode-coupled monostable multivibrator circuits are discussed in detail. Both are practical circuits but one is much more dependent on valve parameters.
The anode voltage/anode current curves used in the designs have $130 \%$ and $50 \%$ of nominal anode current curves added at each relevant value of grid potential, Fig. 1. These are dotted and broken lines, respectively, and are intended to take account of initial high emission and deterioration with life.

Consider first in Fig. 2 the reliability of the multivibrator's stable state ( $\mathrm{V}_{2}$ conducting $\mathrm{V}_{1}$ off). In circuit (1) $\mathrm{V}_{2}$ anode current is controlled by the flow of grid current through R. Thus, without posi-tive-grid data, accurate design is impossible. However, experience indicates that with R in the region of $1 \mathrm{M} \Omega$ the grid/cathode potential will be about 0 V .

Referring to the load line of slope ( $\mathrm{RL}_{\mathbf{2}}+\mathrm{Rk}$ ), the anode current is a nominal 12.8 mA with limits of 14 and 9.5 mA .

Now examine the circuit in Fig. 3. The grid is clamped at -40 V , and since Rk is much greater than 1/gm the cathode will "follow" the grid fairly closely; hence the anode current will be about

40 volts $\div 4700$ ohms $=8.5 \mathrm{~mA}$.
Note that this can be calculated with only a very rough idea of the valve characteristics, thus variations in these characteristics are not going to affect the result greatly. More accurate design using the same characteristics gives a nominal 8.85 mA with limits of 8.95 and 8.65 mA . The method of calcu-


Fig. 1. In the absence of grid-clamping, anode current is seen to vary between 14 mA and 9.5 mA .


Fig. 2. In this cathode-coupled multivibrator, V2 anodecurrent is controlled by an unspecified grid-current in $R$. Accurate design is, therefore, impossible.
lation is given in appendix (1). This considerable improvement illustrates the extent to which tolerances can be designed out by letting the external circuit define the current.

Unfortunately, a price must be paid for this gain. It will be seen that the current in circuit (2) is less but it would appear that by clamping the grid to a higher potential, more current could be passed. However, this is not so. In the low-emission case considered the anode current was 8.6 mA , implying a cathode potential of $8.655 \times 4.7=40.5 \mathrm{~V}$. Thus the grid/cathode potential would be -0.5 V , already at the edge of grid current, so that any increase in grid potential would result in grid current, reducing the accuracy with which the cathode "follows" the grid. If more current is required either more h.t. is needed or the user must accept a shorter valve life.

This reduction in performance is the normal consequence of negative feedback, but in most cases is not too high a price to pay for the simplification in design which has been almost reduced to a problem in Ohms Law. It may be objected that another variable - the clamping-potential tolerance-has been introduced, but the effects of this variable are more easily calculated, and its magnitude more easily controlled than valve tolerances.

Stabilization of the clamping potential may seem attractive, but if this is done, then any variations in the h.t. supply are reflected in full at $V_{2}$ anode, since this valve passes a constant current. Also considerable dependence of pulse width upon h.t. is introduced by stabilization. Consideration of the expression for pulse width:-

$$
\begin{equation*}
\tau \bumpeq C R^{\prime} \log _{e}\left(\frac{\mathrm{E}+\mathrm{V}-\mathrm{Vk}_{2}}{\mathrm{E}+\mathrm{Vco}-\mathrm{Vk}_{1}}\right) \tag{1}
\end{equation*}
$$

shows that if all the terms in the bracket are allowed to vary together there is no change in $\tau$, but obviously this is not so if $\mathrm{Vk}_{1}$ and $\mathrm{Vk}_{2}$ are fixed. Thus the best procedure is to derive all potentials from the same h.t. rail which may be stabilized if necessary.

Also in the second circuit the current in the quasi-stable state is fixed at approximately $20 \mathrm{~V} \div$ $4.7 \mathrm{k} \Omega=4.3 \mathrm{~mA}$ in the same way as in the stable state. Consequently, the value of V is given by $10 \mathrm{k} \Omega \times 4.3 \mathrm{~mA}$ within similar limits. In circuit (2) the magnitude of V has deliberately been kept low so that for a given pulse width the logarithmic term of expression (1) is less than in circuit (1), and C may be increased to compensate giving less dependence on stray capacities. These two points give


Fig. 3. This is a similar circuit to that in Fig. 2, but in this case, grid-current does not flow, and V2 anode current can be accurately determined.


Fig. 4. Negative-feedback and grid-clamping greatly reduce onode current variations. Only a small portion of the curves are shown, in order to show the three " bias lines."
greater control of the pulse width in the second circuit.
It will also be clear that in circuit (1), $\mathbf{R}$ must be high to prevent excessive grid dissipation, without being so large as to give trouble due to negative grid current. In circuit (2), however, the limitations are solely the rating of CR1 and the internal impedance of the clamping voltage. Thus in circuit (2) it is possible to vary the pulse width by variation in R , giving a linear relation, but in circuit (1) E must be varied, giving a non-linear relation.

Enough should have been said to demonstrate that these two circuits, though identical in operation, differ considerably in "designability." Circuit (1) will always show wide variation in performance even when close tolerance components are used. Circuit (2), however, will repay the use of such components, as the variations in performance will be corresponding and simply-calculable, In fact it can be designed on paper with a high degree of accuracy. If the effects of all valve, component and voltage tolerances are calculated the procedure, though simple, becomes a little tedious, even when common sense is used to pick the extreme combinations, but this does save having to "frig" a design at the production stage.

## APPENDIX I

## Calculation of D.C. level in circuit (2)

If $\mathrm{Vgk}=0 \mathrm{~V}$ then $\mathrm{Vk}=40 \mathrm{~V}$ as $\mathrm{Vg}_{2}$ is clamped at +40 . Thus $\mathrm{Ia}=\mathrm{Vk} / \mathrm{Rk}=40 / 4.7=8.5 \mathrm{~mA}$ and this point is plotted on the Vgk $=0$ curve, in Fig. 4.

Similarly if $\mathrm{Vgk}=-2 \mathrm{~V}$ then $\mathrm{Vk}=42 \mathrm{~V}$ and so $\mathrm{Ia}=$ $42 / 4.7=8.9 \mathrm{~mA}$ and this point is plotted on the -2 V curve.

The " bias line" defined by these points cuts the load line at the actual working point. In theory several points are needed to define the bias line as this is not exactly straight, but the error is very small.

By repeating this procedure on the $+30 \%$ and $-50 \%$ sets of curves, the consequential variation in D.C. level can be calculated.

## APPENDIX II

Calculation of Pulse Width.
$\mathrm{Vk}_{2}$ - Cathode potential in stable state.
$\mathrm{Vk}_{1}$ - Cathode potential in quasi-stable state.
V - drop in potential at $\mathrm{V}_{1}$ anode.
$\mathrm{V}_{\mathrm{c} / 0}$ - cut off bias of $\mathrm{V}_{2}$
E , - aiming potential of $\mathrm{V}_{2}$ grid.
$\mathrm{R}^{\prime}-\mathrm{R}+\left[\mathrm{R}_{\mathrm{L}_{1}} \mathrm{R}_{\mathrm{D} 1} \div\left(\mathrm{R}_{\mathrm{L}_{1}}+\mathrm{R}_{\mathrm{D} 1}\right)\right]$
$\mathrm{R}_{\mathrm{D} 1}-\mathrm{D.C}$. . resistance of $\mathrm{V}_{1}$.
$-\frac{(\mathrm{V} b b-\mathrm{V})}{\mathrm{Vk}_{1}} . \mathrm{Rk}$.


Grid 2 potential rises from $\left(\mathrm{Vk}_{2}-\mathrm{V}\right)$ toward E with a time constant $\mathrm{CR}^{\prime}$. Changeover occurs terminating the pulse when the grid rises to $\mathrm{Vk}_{1}-\mathrm{V}_{\mathrm{c} / \mathrm{o}}$.

$$
\begin{aligned}
& \text { i.e. } \mathrm{e}_{g}=\mathrm{E}-\left(\mathrm{E}+\mathrm{V}-\mathrm{Vk}_{2}\right) \mathrm{e}^{-\mathrm{ck}^{\prime}} \\
& \text { at } \mathrm{t}=\tau \\
& \mathrm{e}_{g}=\mathrm{Vk}_{1}-\mathrm{V}_{e / \rho}
\end{aligned}
$$

$$
\begin{aligned}
& \therefore \log _{e}\left\{\begin{array}{l}
\mathrm{E}+\mathrm{V}_{\mathrm{c}} l_{0}-\mathrm{Vk}_{1} \\
\mathrm{E}+\mathrm{V}-\overline{\mathrm{V}} \mathrm{k}_{2}
\end{array}\right\}=\frac{-\tau}{\mathrm{CR}^{\prime}} \\
& \therefore \quad \tau=\mathrm{CR}^{\prime} \log _{e}\left\{\begin{array}{l}
\mathrm{E}+\mathrm{V}-\mathrm{Vk}_{2} \\
\mathrm{E}+\mathrm{V}_{\text {cio }}-\mathrm{Vk}_{1}
\end{array}\right\}
\end{aligned}
$$

## X-ray Image Amplifier

MUCH effort has been directed, in recent years, towards the improvement in sensitivity of X-ray fluoroscopy equipment. Reduction of the X-ray dose incurred by the patient and greater flexibility of operation have been two of the chief aims. By the introduction of the Marconi Instruments 12 -inch Image Amplifier, many former obstacles to the wider adoption of fluoroscopy in diagnosis and treatment have been removed.

Formerly, the light from the fluorescent screen was viewed at the output of an optical system. The disadvantages were that the image was of very low density and that the radiologist, who must be dark-adapted, was obliged to devote a large proportion of his attention to the purely mechanical business of operating the equipment.

In the Marconi system, amplification of the image produced on the fluorescent screen is accomplished elec-
tronically and the image is viewed on a high-definition television-style monitor. The X-ray shadow image is produced on a 12 -inch diameter fluorescent screen and focused by a Schmidt optical system on to the photocathode of a specially developed $4 \frac{1}{2}$-inch image orthicon camera tube. The output is fed, via a low-noise head amplifier situated near the camera tube, to a high-gain video amplifier of $10 \mathrm{Mc} / \mathrm{s}$ bandwidth. It is possible to vary the proportion of the image which is focused on to the target of the camera tube, thereby enabling the operator to control the magnification, while giving greater definition, as the number of lines per unit area of the original image is effectively increased.

The picture on the 17 -inch monitor tube is of 1024line, triple-interlace standard with a horizontal resolution of $10 \mathrm{Mc} / \mathrm{s}$. Cine-recording is provided for and a system known as target-pulsing is employed. With this in operation, alternate fields are stored on the orthicon target, which results in the scan producing twice the normal signal amplitude at half the frequency. The pull-down time of the cine film is thereby not wasted, and during film exposure the picture is twice the normal brightness, which permits the use of fine-grain, slowspeed film. The full potentialities of 1000 -line working are realized, with no trouble from faulty or critical interlace; time-base synchronizing pulses are not mixed with the video signal, as in a television system, but are fed separately.

The video signal polarity may be inverted when copies are required, eliminating the necessity for photographic reversal. It is also possible to reverse the scanning direction, a procedure which is of advantage when the radiologist finds it convenient to reverse the patient on the table.

The most important features of this equipment are the brilliance and convenience of the display, allowing the radiologist to work with no distraction and necessitating no dark-adaptation. The X-ray dose to the patient and, indeed, to the radiologist, is greatly reduced, and several people may view the monitor simultaneously: remotely if necessary.


The 12-in Image Amplifier. The unit containing the optical system and orthicon comera tube is over the centre of the table. The X -ray tube is situated underneath.




## Olympia, London, 30th May - 2nd June







 N.S.F. $\ldots$ Neilf, James, \& Company
Newmarket Transistor


 Permanoid
Plannair

## 



## Elements of Electronic Circuits

26.-Pulse Modulation (1)

By J. M. PETERS, B.Sc. (Eng.), A.M.I.E.E., A.M.Brit.I.R.E.

I$N$ the preceding article (May, 1961) we saw how an open-ended delay line could be used as a pulseforming network in a simple pulse modulator. Other types of pulse modulator are used, however, and some of the various ways of producing a pulsemodulated signal will be dealt with in this and the next instalments. It is usual to divide pulse modulators into two main categories, grid modulators and anode modulators.

Grid modulation, as the name implies, can only be used when the transmitting valve has a control grid, and consequently its use is restricted to frequencies below about $500 \mathrm{Mc} / \mathrm{s}$. At centimetric wavelengths, where the transmitting valve is usually a magnetron, anode modulation must be used.

## Grid Modulation

Fig. 1 illustrates a tuned-anode tuned-grid r.f. oscillator V2 which is prevented from oscillating


Fig. I
by the bias caused by a heavy current flowing through its grid resistor R. This grid resistor forms the anode load of a triode pulse amplifier stage which in the absence of an input is conducting heavily. A negative pulse applied to the pulse amplifier grid cuts off the anode current of V1, produces a positive voltage at V1 anode (i.e., V2 grid) and consequently allows V2 to oscillate for the duration of the pulse. Unfortunately the combination of R, the r.f. bypass capacitor C and the inter-electrode anode/grid capacitance of V1 tends


Fig. 2
to cause a slow rate-of-rise of oscillator grid voltage to zero. This in turn produces a slow build-up of r.f. oscillations and the leading edge of the r.f. pulse is of poor shape. This can be improved by making the value of R small, thus the time constant represented by the total shunt capacitance multiplied by


Fig. 3

R is smaller and the build-up of r.f. oscillations is quicker.
Squegging Oscillator.-Another type of grid modulator is the squegging oscillator which is often triggered by a cathode-follower stage (Fig. 2). First let us consider the action of a squegging oscillator: If the auto-bias CR time constant of the oscillator is too long, a charge is built up on C (by grid current) and the bias voltage becomes greater than the value at which oscillations can be maintained; r.f. oscillations then die away. The charge on C leaks away through R until the bias voltage rises through cut-off when r.f. oscillation recommences. This again causes grid current and an increase in bias so that the cycle repeats. The repetition frequency of these free-running bursts of r.f. oscillation is largely dependent on the time constant CR. If we now trigger the grid circuit of the oscillator, as shown in Fig. 2, the recurrence frequency of the
bursts of r.f. oscillation may be locked to the input (from the sub-modulator). The action of the oscillator is initiated by the triggering pulse which is made longer than the required duration of the r.f. pulse. The duration of the r.f. oscillation is thus still dependent on the rate at which C is charged by grid current flow but, due to the trigger circuit, an r.f. pulse cannot be generated unless a trigger pulse is present. If the duration of the trigger pulse is made longer still, or if the rate-of-rise of mean grid potential after the end of the r.f. pulse is such that cut-off is reached before the end of the trigger pulse, then it is possible for a second r.f. pulse to be produced. This results in the production of double pulses or "doublets" (see Fig. 3). So that the valve is ready for operation by the time each trigger pulse is applied, the free-running repetition frequency is chosen to be greater than the p.r.f. of the trigger pulse.

## IETUTERS TIO THE EDITOR

The Editor does not necessarily endorse the opinions expressed by his correspondents

## Negative Feedback and Non-linearity

ONE must admire the way in which "Cathode Ray" in the April issue handles this very tricky subject, the care and hard work that has obviously gone into the article, and the guarding against snipers at every point, for example "apparent reduction of distortion" in the caption of Fig. 2, and the checking up on what he said in 1946.

Added to all this is his ingenious method of proving that feedback improves the signal-to-distortion ratio, so that it will surely be called mud slinging if I venture to ask if "Cathode Ray" would have trusted this method if he had not checked it experimentally?
I suggest this because hum (or noise) is another important factor in amplifier design, and by substituting "hum" for "distortion" and $h$ for $p$ the method seems to apply practically word for word to hum, in fact even slightly better as in this case the amplifier gain can be assumed constant.
We "prove" that negative feedback increases the signal-to-hum ratio by the feedback factor, but "Cathode Ray" states at the beginning of the article that feedback may or may not increase this ratio, and even may decrease it. The writer believes that it does not change it.

We reluctantly conclude therefore that even this method does not always work, and so, as often happens with the corresponding explanations in text books, we do not know when to be trusting

Another point arises in Fig. 6(a) where it is assumed that $-40,-48$ should read $-32,-40$ respectively.

The maximum negative output possible in Fig. 5 is -24 V , which is assumed equal to a fundamental of $-32 \mathrm{~V}+8 \mathrm{~V}$ second harmonic. This seems fair enough. But it is difficult to understand why it is considered fair to apply an 8 V bias here to make these figures -32 V and -40 V and not to apply any bias in the Fig. 7 case. Binley, nr. Coventry
D. L. CLAY
"CATHODE RAY" in his article on the above topic has raised several interesting points but has, I feel, missed an important one. In discussing the output power available with feedback he has assumed that it is necessary to maintain an input signal centred about the same zero as that without feedback. Since, in theory, the zero value of any signal can be adjusted at
will, it would perhaps be better to consider his amplifier curve with feedback when driven from a $\pm 4$ volt swing centred about the 1.5 volt value of $\mathrm{V}_{1}$. The fundamental output voltage available is then $\pm 40$ volts and is produced with negligible distortion.

If such an arrangement is permissible, and there seems no practicable objection, then some of the conclusions reached in the article would need reconsidering since under the proposed alternative the output power with feedback is certainly no less than that obtainable without feedback.

I do not agree with conclusion 6: large powerhandling capacity in "hi-fi" systems is to enable the occasional peak jevels to be handled without distortion.
Finally his remark about feedback that is not precisely negative is open to doubt since the magnitude of AB and not its phase is important in the reduction of non-linear distortion.
Since the harmonic distortion in a feedback amplifier is reduced in value by the factor $|1+\mathrm{AB}|$ evaluated at the harmonic frequency in a region or frequency range where a reasonable amount of feedback has been applied and the approximation $|1+\mathrm{AB}|=|\mathrm{AB}|$ holds, the harmonic distortion will depend on the magnitude of $A B$ and not the phase of AB .

Dartford, Kent
J. F. TILLY

## The author replies:

I am grateful to Mr. Clay for his appreciative remarks and for pointing out the error in the scale of Fig 6(a). Meanwhile I have observed that the same or another gremlin added noughts to the last two figures in column (2) of the table on the same page.

My statement that negative feedback may either reduce or increase hum was proved both theoretically and experimentally in your issue of May 1946, but as that may be rather far back for even Mr. Clay to recall (though it was reprinted in "Second Thoughts on Radio Theory," Chap. 19) I hope to return to the subject before long, with particular reference to presentday circuit practice.
The question of fairness does not really arise with regard to the 8 V " bias." It represents the d.c. component, which is removed before the signal reaches its ultimate destination.
The comparison made in the article, on which the
conclusions were based, was between Fig. 5 and Fig. 8, the latter representing the same working conditions as the former except for the ten-fold increase of input to make up for the ten-fold reduction of gain due to feedback. It is quite true, as Mr. Tilly points out, that a better result could be obtained-at least on paperwith feedback if the working point were altered. The same means would also achieve a better result without feedback. But it was tacitly assumed that the output stage (which is normally responsible for nearly all the non-linearity) was already being worked at its limit of power dissipation, so that in practice such adjustment would be inadmissible.

It might well be, in any particular practical case, that some slight readjustment of working conditions would be desirable when feedback is adopted, but this does not, I think, substantially affect the main conclusions. The object of my exercise was to bring out what happens when non-linearity at some part of the signal cycle is large-a condition that is inevitable when the input amplitude is excessive.
I am at a loss to understand Mr. Tilly's disagreement with my conclusion 6; his alternative statement is in perfect agreement with it.
There is surely no doubt at all that the phase of the fed-back voltage is important, for it is what decides whether the feedback is negative or positive. In his final paragraph Mr. Tilly commits the logical fallacy of assuming what he is attempting to prove, by putting "magnitude" signs around the feedback impressions he has used.

With regard to the last two words of his penultimate paragraph I would refer Mr. Tilly to the top right-hand corner of p .173 of the April issue.
"CATHODE RAY"

## Further Thoughts on Inductance

THE abstract and intangible quantities involved in the understanding of the behaviour of magnetic circuits seldom fails to present real difficulties to students. It is for this reason that authors must write with great care on such matters.
It is to be regretted that Thomas Roddam writing on
Inductance in the April issue has confused
(a) magnetizing and magnetomotive force
(b) flux and flux density
throughout the whole of his article.
His appreciation of engineers being unwilling to "put back the clock" is very sound. One wonders why the older and more cumbersome magnetic expressions were employed in preference to those of the m.k.s. system.

Finally, no useful purpose whatsoever is served by assigning a proper noun to cover the product or quotient of electrical quantities. On the contrary, a very serious barrier to learning is set up, as the identity of the expres-
sion is lost.

For example, the pedant gives Roddam about half marks for stating the energy stored in a magnetic circuit as $\frac{1}{2} \mathrm{LI}^{\prime}$. "Joules" have been overlooked even as the pedagogue may well have overlooked the fact that his pupils are not very happy about this particular unit and its derivation.

Malvern.

## F. K. WEBB.

AFTER I had tried, in your March issue, to clarify magnetic units and relationships for the beginner, it was a little naughty, was it not, of Thomas Roddam to confuse them again in the April issuc by referring to H as magnetomotive force and $B$ as flux? A lot of readers will be sure enough of their ground not to be tripped up by this loose language, but ought we not to remernber that there are others who are liable to be so confused by such a contradictory usage right at the start of the article as to fall out at that point and thereby fail to benefit from Mr. Roddam's interesting thoughts on non-
linear inductance.

Encouraged by your historical reference, in the same issue, to my efforts to make logic and order prevail in "ur terminology, I will also lodge a protest against "HENRYS" in the sub-title. I know this is in accordance with B.S.560:1934 (Amendment 1), but (owing to my objections?) it was dropped in B.S.1991:Part I:1954 which has superseded it. Clearly the names of units are common nouns, as indicated by their lower-case initial letters, and therefore the plural of "henry," like other English common nouns ending in -ry, is "henries." Only the Welsh could take any pleasure in "rys."
"CATHODE RAY".
The author replies:
Mr. Webb and Mr. Ray are quite correct in reprımanding me for using loose language. Last July, however, and the reference to rain rather than snow indicates that my holiday was in summer, I had not had an opportunity of reading Mr. Ray's clarification.
Why do I get confused over these magnetic terms and why do I not use m.k.s. units? The answer to the second question is easy: our makers of magnetic materials express all their information in oersteds and lines per square centimetre. The answer to the first question is also, I am afraid, an engineer's answer, too. Most of my transformer designing life was associated with a welldefined set of standard sizes of core with a range of standard air-gaps. I think of maximum flux as so many ampere-turns because once a size is chosen that NI term is the only one which is free. Like Mr. Ray (page 145, March issue) I sweep all the dirt under the carpet but I have chosen a different corner.

I must disagree with Mr. Webb's blanket damnation of proper nouns for products and quotients. Does he really want to use volts per amp for his resistors and kilovolt-amps for his electric fire?

Finally, sir, should I be taken to task? When I was a student I was assured that it was my stupidity which led me into those difficulties. Surely either the students of today or other pedagogues of yesterday are to blame. The manufacturers of magnetic materials could, however, make it easier for us to fall into righteousness.

THOMAS RODDAM.

## Graphical Solutions

MAY I suggest that the graphical solution proposed in the Appendix to Foss and Sizmur's article in "Multivibrator Design" (p. 224, April issue) would be made even simpler by the use of linear/log scales?
To provide a graphical solution which really replaces the use of (low-accuracy) tables, $\mathrm{E}_{1} / \mathrm{E}_{0}=\mathrm{e}^{-\mathrm{t}, \mathrm{T}}$ should be drawn on linear $X /$ logarithmic $Y$ paper, whereby it becomes a straight line, a property which this function shares with all functions of the form $a^{x}$. It is then only necessary to know one point on this graph in addition to the point $(0-1)$.
As an example a straight line was ruled from $(0-1)$ to ( $4.60-0.01$ ) on paper with two log decades along the $Y$-axis and a linear $X$-axis scale. The accurate figure for Y at this point, taken from "Tafels van $\mathrm{e}^{x}$,"
H. W. Holtappel (Groningen, 1938), is 0.0100518357 . These tables show that the point (4.605-0.010) is closer to the correct figure, but the easily memorized co-ordinates (4.60-0.01) give ample accuracy for the purpose at hand.

The above assumes that linear-log graph paper is available, but in an emergency the two decades of the vertical scale can be transferred from the upper scale of a slide rule. If an ordinary $\log$ table can be assumed to be available, the ordinates can be marked off in proportion to the logarithms of $1,2,3$, etc., the units being centimetres, inches or multiples thereof.

Finally, when required, it is easy to draw the curved graph on ordinary linear-X/linear-Y paper, taking individual points from the straight line graph first
prepared. prepared.
Ramstad, Norway. KAYE WEEDON.

The Marconi Instruments TFI 345 lOMc/s Counter, showing the plug-in, printed-circuit decade units.

By P. R. DARRINGTON



## DIGITAL MEASUREMENTS

MPROVEMENTS in the stability of oscillators have brought in their train a demand for more and more accurate methods of frequency measurement. Accuracies of a few parts in a hundred or even a thousand million are becoming commonplace, and the most easily operated instruments for measurements of this order are the digital frequency-meters now being produced in many parts of the world.

Digital measurements of time and the counting


Fig. I. Dekatron.
of random events have been adopted by the industries concerned with nuclear energy, and many types of input transducer have extended the advantages of digital measurement to the engineering industry in general.

The advantages of this type of measurement may be summed up as extremely high accuracy, simplicity and freedom from operator fatigue, the latter being influenced to some extent by the type of read-out presentation adopted.

Counters generally available are divisible broadly into two groups. The lower-speed instruments tend to rely on the counting tubes, as instanced by the Dekatron. This is a gas-filled device consisting of a central anode surrounded by ten cathodes (K0 to K9 in Fig. 1). Between each pair of cathodes are two electrodes known as switching-grids. On first switching on, an ionization path is formed between the positive anode and one of the cathodes which are returned to earth: from this time, the path will be formed to whichever is the most negative electrode in the tube. If a negative-going pulse is applied to all the grids connected to $S 1$, the path will move to the S1 grid nearest the original cathode; it cannot move backwards to S 2 , as S 1 is more negative. A second negative-going pulse, delayed slightly to overlap the first, is now applied to S 2 . When the S 1 pulse ends, S 2 is more negative and the path is formed here. At the expiry of this second pulse, the path moves on to K1, as all cathodes are maintained at a more negative d.c. potential than the grids. The two pulses are obtained from a mono-
stable multivibrator triggered by the input signal, a delay network providing the overlap. Ten pulses at the input to the multivibrator result in the ionized path performing a complete circuit of the tube. A resistor in K0 will develop a voltage when the path is formed between K0 and the anode, which results in division by ten. Reset is effected by the momentary application of a negative voltage to K 0 . The position of the path is identified by the formation of a glow on the cathode in use, which corresponds to a numbered surround on the instrument front-panel. This type of presentation can be fatiguing to the operator, and the principal disadvantage is that the maximum switching frequency is limited to about $20 \mathrm{kc} / \mathrm{s}$. Counting tubes are used widely in comparatively low-speed industrial applications.
A device belonging to the family of counting tubes, but working on a different principle is the beam-switching tube. This is a vacuum tube, the electron stream being directed from a central cathode to one of ten anodes, or "targets," by the combined influence of a magnetic field and the voltage applied to a grid or "spade." The maximum switching speed of later developments of the beam-switching tube is in the region of $10 \mathrm{Mc} / \mathrm{s}$, and the tube will drive read-out devices, such as neon tubes and printers directly.
Most of the higher-speed counters- $10 \mathrm{Mc} / \mathrm{s}$ and above-employ as their basic counting element the bi-stable flip-flop or toggle. Circuits using tran-
sistors are capable of switching at speeds in excess of $50 \mathrm{Mc} / \mathrm{s}$, although in commercial instruments the present-day limit is about $20 \mathrm{Mc} / \mathrm{s}$.

## Principle of Operation

The basic requirement of an electronic counter, when used other than purely as a batch counter, is that it should have an accurate and stable source of frequency. When built into the instrument, this almost invariably takes the form of a temperaturecontrolled crystal oscillator, usually working at a frequency of $100 \mathrm{kc} / \mathrm{s}, 1 \mathrm{Mc} / \mathrm{s}$ or $5 \mathrm{Mc} / \mathrm{s}$. The $5 \mathrm{Mc} / \mathrm{s}$ crystal is being widely adopted because of its low long-term ageing drift. The temperature of the crystal, and sometimes of the complete oscillator is held constant by enclosing it in an oven, controlled by a bimetallic thermostat, or an electronic circuit using a temperature-sensitive element in a bridge circuit. The more sophisticated control circuits are capable of working with differentials of a few millidegrees; divergence from nominal frequency is normally limited to not more than a few parts in $10^{7}$ per week, or even an order better. Oven temperature control employing the latent heat of fusion has also been described.*

The output of the crystal oscillator is used as a reference frequency in the gating and timing cir-
*Wireless World, Dec. 1960, p. 613, "Technical Notebook".


Fig. 2. Circuit configurations for frequency and time measurements.


Fig. 3. The $\pm 1$ digit error inherent in the system of gating is due to the lack of phose relationship between the crystai oscillator and the unknown frequency.
cuits of the counter. Two main configurations of the internal circuits are required, providing the two functions of frequency and time measurement. Block diagrams are shown in Fig. 2.

The obvious method of frequency measurement is to allow the input to pass through an electronic gate which is open for one second precisely. The output of the gate is fed to a chain of counting decades of one of the types described earlier; the result showing on the display when the gate closes is the frequency in cycles per second. The accuracy of the result is governed by the stability of the crystal oscillator which indirectly determines the opening time of the gate, and also by an inherent error of $\pm$ one digit, caused by the lack of a phase relationship between the frequency being measured and the output of the crystal oscillator (Fig. 3).

At very low frequencies, say $10 \mathrm{c} / \mathrm{s}$, the accuracy is determined by the gating error (ic/s) which is, in this case, $\pm 10 \%$. The method of measurement is therefore inverted. In order to obtain a larger number of significant figures on the display, the unknown frequency itself is made to determine the gate opening time, during which a high frequency derived from the crystal oscillator is counted. In this way, the gating error is reduced, the major inaccuracy being due to crystal frequency tolerance, which is controllable. The result displayed is now the time of one cycle of the unknown frequency in units of the standard frequency being counted, and the frequency is obtained by a reference to a set of reciprocal tables.

Measurement of time is made in exactly the same way; the input determines the opening time of the gate, this being defined precisely by the standard fi Tuency derived from the crystal oscillator. In some counters it is possible to vary, by means of a calibrated front-panel control, the standing bias on the trigger circuit which controls the gate opening time. The points on a waveform at which the gate opens and closes are therefore variable. This is a useful facility for the measurement of pulse widths at different amplitudes, the time taken for a slowlyvarying voltage to cross predetermined levels, etc.; but it must be borne in mind that the accuracy of measurement has a limit set by the maximum frequency employed as a standard. For instance, if the standard frequency used is $10 \mathrm{Mc} / \mathrm{s}$, the accuracy cannot be better than $\pm 0.1 \mu \mathrm{sec}$. In many counters, automatic positioning of a decimal-point gives the result in $\mathrm{kc} / \mathrm{s}$, or, if time is bcing measured, in the appropriate timing-units.

The measurement of frequencies outside the range of the basic counter is made possible by recourse to heterodyne techniques. The method is to beat the unknown frequency against a multiple of the crystal frequency, selected by a front-panel control, and obtained by harmonic generation or locked-oscillator
techniques, the difference frequency being measured on the counter. The presentation is then the selected multiple added to the counter indication. For instance, if the frequency to be measured is $185.734 \mathrm{Mc} / \mathrm{s}$, the harmonic selection control is set to $180 \mathrm{Mc} / \mathrm{s}$. The difference frequency is fed to the counter and is displayed as $5734 \mathrm{kc} / \mathrm{s}$. The problem of which harmonic to select is usually resolved by means of a wavemeter. This method is applicable up to about $200-300 \mathrm{Mc} / \mathrm{s}$, above which it is general practice to employ a transfer-oscillator. This consists of an oscillator operating within the range of the counter, with or without heterodyne facilities, harmonics of which beat with the unknown frequency. Provided that the harmonic is known, the result is obtained by measuring the frequency of the transferoscillator and multiplying by the harmonic number.

It has been mentioned that the crystal oscillator determines the opening time of the gate. To achieve this, the crystal oscillator output must be divided in frequency to give pulses spaced in time by an


Fig. 4. A typical phantastron divider. The division ratio is set by the value of $C$ and $R$.
amount equal to the gate time required. In practice, several frequencies are produced to give a selection of gate times depending on the frequency being measured and the accuracy required.

Circuits employed as dividers are well-established, among them being blocking oscillators', phantastrons ${ }^{1}$, multivibrators ${ }^{1}$ and the diode-pump integrator with transistron linearization and flyback ${ }^{2}$. Fig. 4 shows a typical phantastron divider. Characteristics required are stability with temperature and supply voltage changes and ease of setting-up. The divider should not free-run in the absence of input pulses.

All these circuits depend on a time-constant, which, at frequencies below about $10 \mathrm{c} / \mathrm{s}$, becomes


Fig. 5 (a). A counting decade, showing the signal and feedback paths in heavy lines. (b) Waveforms appearing at the anodes of valves whose grids have the reset pulse applied.
unmanageably large, and insulation resistance of circuit components tends to cause trouble. To overcome this, aperiodic circuits, such as toggle decades are used, there being no lower limit to their frequency of operation. Many transistor circuits use decades throughout the dividing chain.

A self-checking facility is incorporated in most counters which tests for correct working of all division, gating and counting circuits. The gating circuitry is set to function as in the case of fre-quency-measurement, while the output of the crystal oscillator and the stages in the division chain are counted.
The result should be exactly correct, although accumulated phase-shifts in the division chain, and the finite opening and closing times of the gate circuit may result in a $\pm 1$ digit error.
This check does not give an indication of crystal accuracy, as a low crystal frequency gives rise to a longer gate time, and the two effects cancel out.

## Decade Operation

Referring to Fig. 5a it is seen that the decade is composed of four bi-stable toggles, the output of each triggering the succeeding stage: the last toggle provides the input to the next decade.
The line marked RESET may be considered, for the moment, as being returned to earth. Assume an initial state where all left-hand valves are conducting.
The grid of a toggle in the conducting condition is returned to a positive voltage with respect to
cathode, but is held at very nearly cathode voltage by grid current limiting. When cut off, the grid may be up to 50 V negative. Obviously, then, a positivegoing pulse applied to a toggle will not affect the conducting valve, and will need to exceed 50 V to trigger the stage. On the other hand, a small nega-tive-going pulse will be amplified by the conducting valve, and will "flip" the toggle into the reverse condition. All trigger pulses, therefore, are negativegoing.

The first input puise triggers toggle 1 into the condition where the output anode (left-hand, has performed a positive-going step-function. The step is differentiated by CR, but this has no effect on toggle 2 for the reason explained above.
The second pulse compels toggle 1 to revert to its original state, the negative-going step-function on its anode being differentiated to form a negativegoing pulse suitable for triggering toggle 2. Toggle 1 continues in this manner; a negative-going output pulse is delivered for two pulses at the input.
At the onset of the fourth pulse, the output of the second toggle is of the correct polarity to trigger toggle 3, which is duly reversed. However, at the opposite anode of toggle 3 , a negative-going stepfunction is performed which is differentiated by C R and used to re-trigger the second stage. In the absence of feedback this state of affairs would not be achieved until the sixth input pulse. The count has therefore been advanced by two.
Operation proceeds normally until the sixth pulse, when feedback occurs between toggle 4 and toggle 3 . The "natural" count of $2^{\prime}$-sixteen-has been


The Racal Type SA505 10Mc/s Digital Frequency Meter. Transistor decades are used in both the counting and dividing chains.
reduced to ten by feedback, and the tenth pulse restores all toggles to the initial condition.

The system of feedback is typical of operation at frequencies up to about $100 \mathrm{kc} / \mathrm{s}$. At higher frequencies, although the principle remains, accumulated delay in successive stages of the decade enforces refinements such as gated feedback, whereby the first toggle triggers the fourth directly, via a gate. Heavy clamping of anode and grid excursions is required to reduce recovery times and valve ageing effects.
Reset is effected by applying a large-amplitude positive-going pulse to all the left-hand grids. The source impedance of the circuit producing the pulse is low, to avoid grid impedance unbalance between the toggle halves.

The information stored in the decades is indicated by the condition of each of the toggles when the gate closes, and is readable by several different methods. The most often used are meters, graduated $0-9$, vertical decades of neon lamps numbered $0-9$, and several kinds of in-line read-out, consisting of a horizontal row of numerals, one per decade.

A meter display is probably the simplest to provide, as it consists of only four resistors and a meter. Each toggle has connected to one of its anodes a resistor, the four resistors being taken to a common output which feeds the meter, as in Fig. 6. The problem is to assign to the four resistors values which will give a current output proportional to the significance of each toggle. In a "decade" without feedback, these four currents would increase in powers of two, i.e., the first toggle would signify 1 , the second 2 , the third 4 and the last one 8 . This
is known as pure binary code and is often referred to as 8421 code. However, with the addition of feedback necessary to make the four toggles work as a decade, the significance of each toggle is altered; the decade shown in Fig. 5a produces a 4221 code. If each resistor is given a value calculated to provide the required current, the result will be a tenstep staircase. The system described is used in several commercial counters, but suffers from the disadvantages common to analogue devices presenting digital information. The meter movements are comparatively slow in operation, and in low-speed decades, where anode travel is not clamped, varying valve characteristics necessitate setting-up of limits, by means of variable resistors. Considerable effort is required to make a long series of readings; the result is not immediately apparent.

Read-out employing vertical decades of neon lamps is used widely and is probably the ideal form of display. Each neon lamp is fed via resistors from the anodes of three toggles, the connections being arranged so that one cut off and two conducting valves are connected across the appropriate neon for each indication. This avoids any tendency for more than one neon to fire at any one time. This form of display is instantaneous in operation and has the advantage that it is, in effect, a posi-tional-analogue indicator, in addition to being digital in form. The point is of some importance on many counters, as the display and reset time can be so short as to give a virtually continuous reading. Variations of the input parameter are apparent as a slow rise or fall of the decade indications, and the direction of drift of frequency or time is obvious.

One method of in-line read-out is an extension of the method described above. In front of each neon lamp of the vertical decade is mounted a photo-electric cell. As the neon lights, a voltage is applied to the appropriate electrode of a gas-filled indicator tube. Each electrode consists of a wire formed to represent a numeral $0-9$; as the decade neon lights, a glow forms on the indicator tube electrode.

Another common method of obtaining an in-line read-out is to decode the four outputs of a counting decade by means of either semiconductors or relays, Fig. 7. A conducting path is formed from the common input of the decoding matrix to one of ten outputs by the relay-controlling valves or transistors, which are controlled by the four decade outputs. Current is fed via this path to one of a series of ten lamps in a projection display unit.


Fig. 6. Circuit giving meter readout. The anode resistors are graded to accommodate the differing "significance " of each toggle.


Fig. 7. A relay decoding matrix to convert a 4221 code to a decimal output.

Each lamp has a focusing-lens and a number mask, the number being projected on to a ground-glass screen.

The meter read-out referred to previously is capable of modification to provide an in-line display. The movement carries, in place of the needle, a transparency, consisting of the numerals 0-9 and moving in an optical system. The appropriate numeral is thereby projected on to a screen. The system is especially useful as an indicator for use with transistor decades, where voltages of a magnitude suitable for the operation of neons do not exist.

In-line displays are most useful for a measurement in which the reading is either used once only, or stays sensibly constant during a series of readings. If the input is varying, the display becomes a mere jumble of numbers and cannot be read if the cycling time of the counter is short. The analogue feature of the vertical display is then lost.
Both the vertical display and the in-line read-out may be read from a distance; the vertical display by virtue of the fact that the numbers do not have to be read, as the position gives the result, and the in-line system because the numbers can be made large.

The type of equipment so far described is used, as has been said, for the measurement of frequency and time. An extension of the principle is that of voltage measurement. The direct voltage to be determined must first be converted into either a frequency or a time-delay. A common method is to trigger a time-delay circuit, e.g., a phantastron, by means of the pulse that also opens the gate. As the phantastron anode "runs down," a comparator compares the anode voltage with the amplified input voltage. When the two are in coincidence, a pulse is formed which closes the gate. During the gate opening time, pulses from a crystal oscillator have been counted by a series of counting decades, and the displayed result can be arranged to indicate voltage numerically.

Errors in this method of measurement stem from non-linearity of the time-delay circuit, drift in the direct-coupled input amplifier and possibly frequency error in the circuit providing the standard frequency pulses.

An alternative method is to control, with the amplified input, the frequency of a pulse generator, the output of which is then measured as in the case of a normal frequency measurement. In this case, the inaccuracy is due to a non-linear relationship between voltage and frequency in the pulse generator and drift in the d.c. amplifier.

To obtain a more linear voltage-frequency conversion, especially at the extreme lower end of the range, a common method is to employ a b.f.o. system, in which one oscillator is made voltage-variable. In this way, the transfer characteristic may be made linear down to zero frequency.
It frequently becomes desirable to preserve the results of a serics of measurements in the form of a numerical printed record. To achieve this, some form of decoding is necessary between the binaryform of the decade four-wire outputs and the required decimal presentation. This may be achieved either electrically or electromechanically. The most common method is to employ a relay matrix, shown in Fig. 7, controlled by valves which have on their grids the outputs from each toggle of a decade. The relay contacts are arranged to provide, for each decade indication, a path from the input of the matrix to one of ten outputs, which fecd a printing machine.

An analogue output which may be used to drive a pen recorder is easily obtained from a resistive network on the four outputs as in the meter display circuit. If several decades are in use, the analogue outputs from each decade may be combined in a further network to provide, for instance, units, tens and hundreds.
It can be seen that digital techniques afford, in general, a convenient and rapid method of frequency and time measurement with a high potential accuracy. With the advent of v.h.f. transistors and tunnel diodes, it is inevitable that counting-speeds will rise and that the simplicity of operation and compactness of the counter will bring measurement at an accuracy of 1 part in $10^{7}$ down to the level of the production line.

## REFERENCES

${ }^{1}$ A. H. Frederick, et al. "Waveforms." M.I.T. series, Chap. 16.
${ }^{2}$ P. R. Darrington. "Marconi Instrumentation." Vol. 7, No. 3. Sept., 1959. Page 82.


This closed-circuit television camera, made by Marconi's W.T. Co. Ltd., is only $2 \frac{5}{8}$ in diameter by $15 i n$ long. It can thus be employed conveniently for the internal inspection of pipes and bore holes. The control unit is also shown.

# Leakage in Printed Circuits 

## PROS AND CONS OF WAX COATING

AS A CURE

By P. RUSHEN*

THIS article is based on some original work done on printed wiring and its use in conditions of high humidity and temperature. Cost of materials and ease of production were two of the factors considered throughout, whilst the successful use of printed wiring in exported radio receivers was the main object.

## Effects of Leakage

Consider some aspects of a conventional valve circuit using printed wiring-in general, every valvebase tag will be in intimate contact with the base material and in some cases anode and grid tags are adjacent to each other. An a.g.c. line may have as


This drowing, extracted from B.S. 1137:1949, $\dagger$ shows the recommended arrangement of specimen and electrodes for test of insulation resistance. The specimen is immersed in water for 24 hours at $20^{\circ} \mathrm{C}$ (Type 1 material) or subjected to relative humidity $75 \%$ at $20^{\circ} \mathrm{C}$ for 48 hours (Type III material) before electrical test is applied.
many as four or five contacting points. The h.t. line, the filament supply and, possibly, near-mains voltages all make close contact with the base material at various points on a typical panel.
The effect of any leakage resistance will of course depend on the type of equipment concerned: an unrequired component at a significant point in a

[^1]

Test piece, showing arrangements used by author for determining surface leakage on material (left) and (right) insulation resistance of "body" of board. In latter case pins are driven through board: consequently surface contact is negligible.
radio or television receiver can cause malfunction and, possibly, complete failure. A $100-\mathrm{M} \Omega$ leakage resistance from a $6.3-\mathrm{V}$ filament line to the control grid of an amplifier could result in an audio power output of 100 mW or more. No less depressing may he the prospect of a similar value of resistance from the h.t. line to an a.g.c. point.

## Base Materials

If one were able to consider the use of only the more expensive base materials, the foil-to-foil leakage resistance on the surface would be the main factor. For practical purposes, however, one is obliged to look to the less expensive and more common syn-thetic-resin-bonded paper (s.r.b.p.) materials and those considered here are in general use and conform to British Standard Specifications 1137 Type 1, 2572 Type Pl and Type P3. When used as a base material for printed wiring, the insulation resistance of these materials becomes as important as the surface leakage resistance.
With these base materials one could therefore anticipate:-

1. Leakage through the base material.
2. Foil-to-foil leakage through any surface material whether intentionally added or otherwise acquired.

## Base-material Leakage

Insulation resistance can be related fairly well to the quality of the material and thus, inevitably, to the cost, for which $25 \%$ of the total cost of a board, printed, etched, punched and finished appears to be acceptable. Although the quality of the material undoubtedly has some effect on the punching operation, it seems unlikely that the cost of the complete process would be very much affected. For example, an increase in the cost of base material of, say, $20 \%$ would probably result in an increase of $5 \%$ in the cost of the finished article.

The British Standard Specification for the three materials considered above calls for a 24 -hour period of water immersion prior to measurements of insulation resistance. We might be inclined to regard water immersion as rather a hard test and the re-
sults as a little unrealistic; but after subjecting samples of each material to exposure at $35^{\circ} \mathrm{C}$ and $95 \%$ relative humidity, leakages far worse than those specified for insulation resistance can be obtained over similar distances in the material. Deterioration in insulation resistance is gradual and continuous with hours of exposure; recovery is also a slow process. The insulation resistance of the base material is mainly dependent on the average conditions of humidity. Tests on the three materials considered led to the conclusion that only the best quality (B.S.S. 1137/1) would be free of serious leakage resistance after only 48 hours of tropical exposure.

## Surface Leakage

Evidence suggests that the surface leakage is not affected seriously by the quality of the base material but is dependent far more on the nature and quality of surface accretions. These may be of any substance which covers the board surface and may include chemicals left by the etching process or subsequently acquired, any fluxes, waxes or preservatives which are added to or allowed to remain on the surface of the board. The volume of surface material will in all cases be small compared to the volume of base material and this gives the surface leakage a rather different character. The insulation resistance of the base material is "solid" and semi-permanent, the surface leakage is variable by comparison. The variability of the surface leakage is usually due to the absorption and evaporation of moisture from the surface material. If the volume of surface material is small, it can be expected to dry out very rapidly with any increase in temperature or any improvement in local conditions of humidity. Power dissipation in the surface material due to leakage current, although very small, assists the evaporation of moisture. In mains-operated or heated equipments, leakage resistance in the surface material may well disappear before its effects become apparent. Unfortunately in practice we cannot separate the two forms of leakage resistance, neither can we, being aware of the effects of time and environment, be very sure of the quality or quantity of the surface material.

## Improvement of Existing Materials

We are obliged to consider whether we can improve the conventional materials. The possibility of adding slots in the board was first considered; but this exposes more of the inner material and thus leads to a greater absorption of moisture and worsened leakage.
The most successful results were obtained with a wax treatment of the material. Samples of base materials were immersed in a wax bath at a temperature of $130-150^{\circ} \mathrm{C}$ for 3 to 5 minutes and a
considerable improvement in insulation resistance was obtained. Sample boards, etched and pierced but without components or fixtures, were treated in a similar fashion and the results were equally successful. Even the lowest grade of material gave surprisingly good results when treated and maintained its improved qualities after 200 hours of exposure.

The semi-absorbent nature of the inner material, assisted by some de-lamination which occurs as a result of the punching operation, permits absorption of wax at every hole or edge on the board. The thickness of wax deposited on the board is about .001 ". Unfortunately one cannot achieve the same degree of success with wax treatment applied after assembly, because the holes which would provide access to the inner material will be sealed with solder. Some success may be achieved by having open and unoccupied holes or slots at known vulnerable points but an immersion time of at least three minutes may well be impractical with a complete assembly. A "flash" dip of assembled board has very limited value by itself.

## Pre-assembly Waxing

The pre-assembly waxing process, though adequately fulfilling requirements for insulation, has a number of disadvantages. Whether these are acceptable will depend on individual considerations of quantity, production facilities and various other factors, not the least important of which will be the standard of quality acceptable for the product. Many manufacturers are using hand soldering processes on printed wiring in quantity production and reduced reject rates and greater reliability are achieved in most cases. The wax treatment of panels to be hand soldered has little effect on the soldering process but the "people who do the work" are sensitive to changes of any sort and some difference will be apparent to them.
The wax treatment may render difficult the utilization of automatic soldering processes. Individual evaluation would probably be necessary for a specific assembly and process.

Although it is normal to complete the board production process with an application of fluxing varnish of some sort, the waxing process makes this impractical and unnecessary. It becomes essential, therefore, that the waxing process immediately
follows the post-etching cleaning and drying. It replaces the lacquering process both in time taken and sequence of operations.
It is possible that long periods of storage of unassembled boards may make a subsequent soldering operation a little difficult but it is to be emphasized here that the protection afforded by a surface preservative is also somewhat limited. Some unlacquered waxed boards were exposed to extreme conditions of humidity and others left outdoors, in both cases for some days, without any apparent deterioration in solderability.

## Mounting of Panels

It is possible that some edges of the board may have excess wax deposits due to the drain-off after removal from the bath. If by misfortune these occur
at mounting points, it may be necessary to remove the excess to establish a firm fixing. It is possible of course that arrangements can be made for the board to be drained in such a way that any surplus wax does not build up on the mounting positions.

## Conclusion

The pre-assembly waxing should enable use to be made of the lower grades of base material. It may introduce some minor difficulties and inconveniences into a production system but these would appear to be a small price to pay for the degree of success likely to be achieved. The usual care and attention paid to the cleaning and preparation of the surface and the common-sense arranging of foils and holes to isolate vulnerable points in the circuit from high potentials are, of course, still necessary.

# TELEVISION AND FILM TECHNIQUES 

HIGHLIGHTS AND POINTERS FROM A RECENT CONVENTION

PENNY plain and tuppence coloured" was definitely the tenor of the convention on television and film techniques held at the Institution of Electrical Engineers on 20th and 21st April, sponsored by the British Kinematograph and Television Socictics. Of the eight papers, five were primarily concerned with colour: one other paper contained more than a passing mention, and delegates insisted on raising the subject in the discussion following the two others.

Naturally, great general interest centres on the choice of standards for colour television, and those who favoured an adaptation of the N.T.S.C. system (which, it will be remembered, uses phase and amplitude modulation of a subcarrier to transmit colour information and a.m. of the main carrier for luminance or "brightness") might well have had doubts raised in their minds by the paper and demonstration by Polonsky, Sauvenet and Cassagnc. The system was the Henri de France* (or as it is now much less euphoniously known-SECAM) in its latest form. The original system used a.m. of the subcarrier to transmit colour-difference signals, the two signals required being transmitted sequentially on alternate lines of the picture. Changing of the subcarrier modulation to f.m. has resulted, in the main, in a reduction of sensitivity to the effects of noise. Visibility of the f.m. subcarrier has been reduced to the point where, at the moment, it is only 4 to 5 dB worse than in the N.T.S.C. system. That the performance of SECAM in the presence of transmission distortions such as phase errors or differential gain effects is vastly superior to N.T.S.C. was demonstrated by the switching-in of such distortions and noise. In the case of a delayed secondary signal (ghost) the amplitude and delay was such that the line synchronizing of one receiver was upset, but no effect on the colour could be seen. It was claimed that the stability of the system was good enough to eliminate the need for either hue or saturation controls; although it was stated by Rodgers (Bush) in discussion that experimental 405 -line N.T.S.C. receivers have been made without such extra "colour" controls. A feature of the discussion was the listing, by Maurice (B.B.C.), of points for and against SECAM in comparison with N.T.S.C.: the results were close, being slightly in favour of one system or the other depending on the viewpoint
taken for assessment. It is perhaps appropriate at this point to note the gist of Polonsky's reply to another questioner: over the years thousands have worked on N.T.S.C., SECAM has occupied ten or fifteen people only for a shorter time.

Pictures for the demonstrations were generated by a flying spot scanner at the Hirst Research Centre of G.E.C. and transmitted over radio link and cable.

## Colour on Record

The recording on magnetic tape of N.T.S.C. signals was the subject dealt with by Roizen (Ampex International) who reviewed the methods employed and described a new technique, which (as he pointed out) like all good solutions to problems, is extremely simple. The major difficulty is that an inaccuracy in phase greater than $5^{\circ}$ is unacceptable, and timing has to be held accurate to $0.002 \mu \mathrm{sec}$. The scheme that was put into operation some time ago is known as "burst lock." On the American standard, about eighteen lines are recorded in one sweep of one recording head: within one line the errors caused by non-uniform head motion could not, due to the inertia of the head system, exceed the permissible limits. Thus by taking the subcarrier reference burst or pulse which occurs during the back porch of each line-sync pulse and "remembering" its phase with a ringing oscillator correct relative information of colour can be achieved for that line. Then this can be translated back to the stable subcarrier reference required for transmission. This must, as is apparent even from the over-simplification above, lead to much "electronic" complication. The new method, though, is very much simpler and was developed from a device called Amtec (AMpex Time Element Compensator). Amtec was designed to remove the "cogging" effect sometimes apparent on recorded vertical edges when a display employing a flywheel line timebase is used. The device uses a lumped-constant delay line whose capacitors are back-biased junction diodest. Alteration of the bias, and thus the capacitance, provides a continuously variable delay to which the vision signal is subjected. The sync pulses from the tape replay are compared with a source of local, stable sync: when a timing

[^2][^3]error develops, the error voltage is made to change the delay time to compress or stretch out the picture line to its right length, so correcting the "cogging." For colour, the process is carried a step further by comparing, after time correction, the phases of the replayed and local colour burst to develop a "fine " correcting signal.

Another method of registering video signals, in this case by the use of a thermoplastic recording technique $\ddagger$ was described by Glenn of the General Electric Research Laboratories (U.S.A.). This technique employs a base material of relatively high melting point coated with a thermoplastic film which is softened by r.f. heating. A "picture" is written in the form of surface charges on to the material by an electron beam: the film is then softened and the electrostatic forces caused by the charges form "ripples" which can be fixed by cooling the film below its melting point. Visual reproduction can be achieved with an interference-type optical system giving a picture which can either be displayed on a screen directly or converted to a television signal for transmission. The demonstration used optical projection: a grating is imaged by a lens on to another grating so that the passage of light is blocked. The ripples of the recording diffract the light allowing it to pass through the grating to be focused on the screen.

Colour recordings can be made too: for this the image itself is registered as very fine diffraction gratings. A photograph of a reproduced thermoplastic recording of one of the "standard" test slides was shown.

## Colour Techniques

One source of input to a colour television system is likely to be colour film. Grimshaw (British Kodak) read a paper by Veal and Ritchey of Eastman Kodak, in which they described the measures they had found to be advisable in the production of colour films for television. Measurements of the overall transfer characteristic from a flying spot or three-vidicon teleciné equipment to a display using the shadow-mask tube were made. One test, for contrast range, was accomplished by taking an "average" scenc and punching holes in the transparency. The holes were then covered by known densities to discover, against the background of an average picture, the highlight and "black" shadow levels usable. Other tests discovered the definition realizable, colour rendering and noise level corresponding to grain in the film. Broadly the conclusions drawn were that contrast ratio should not exceed $40: 1$-corresponding to a key light/ general light ratio of about $2: 1$-and that the sharpness and grain acceptable for the U.S. system could be achieved by 16 mm film. Colour rendering capabilities of film and TV are slightly different and, in fact, television covers a greater overall area on the colour triangle.

It appeared from the discussion on this paper that the European television engineer would far rather achieve the down-grading of film's full contrast range by electronic means: then as the system was improved the full potential of the film could be employed. The reduction in contrast range necessary would be achieved by a luminance feedback "compression" system: if this were employed with an improved TV system to "restore" a tilm specially produced for television, it would have to be a type of positive feedback and thus would be difficult to control.
Davies (Kodak) described a compact new machine for the fas: processing of black-and-white film for television: the technique makes use of viscous solutions applied to the film.

Stanley and Treays (B.B.C.) gave a thorough account of the experimental programmes broadcast by the B.B.C. since 1954. The purpose of these tests was to assess the $405-$ line version of the N.T.S.C. system and also to discover its artistic possibilities. Many of the techniques for studio productions turned out, in fact, to be similar to the techniques for using colour film, although, judging from the account, some of the "subjective" effects of

[^4]colour (the influence of the appearance of one colour by another) may well prove rather more serious under television conditions. This paper was followed by Stanley and Watson (B.B.C.) reviewing some of the equipment problems and techniques, illustrated by N.T.S.C. colour signals from Lime Grove thrown on a $12 \times 9 \mathrm{ft}$ screen by a new Marconi projector.

One objectionable form of distortion is that which results from gain inequalities at different signal levelsin a black-and-white system this distorts the relative contrast between tones but normally is not obvious. The effect on a colour picture of a change of gamma from 0.6 to 0.4 (demonstrated by Stanley and Watson) was to alter a peaches-and-cream complexion to florid, almost fevered appearance! Shading signals, which have to be applied to enable a camera to reproduce an even field of illumination, can result in spurious colours in patches of low luminance. Great care has to be taken here and one solution might be to "common" the red, green and blue signals at low levels. Phase distortion, too, is most important-a shift of $5^{\circ}$ in subcarrier phase can produce objectionable colour changes; also the behaviour of a receiver's detector when presented with components in quadrature can, with positive vision modulation, reduce the saturation of colours. However, the fact that it is possible to make a system which produces good results was demonstrated by the switching into the link carrying the demonstration picture of two low-power transmitters and two receivers, all in series. The major effects (and the large picture was being viewed from only thirty feet away) were a slight reduction in saturation and a rise in noise level.

## Large-screen Projectors

The slides used in the demonstration were followed by two films which, due to their content, must have placed severe demands on the system and the projector which, incidentally, used three English Electric c.r.t.s with Schmidt optical systems. Beam current was 1 mA maximum for each tube (at 50 kV e.h.t.) and a screen brightness of about 5 foot lamberts was realized. The definition achieved was a great advance on earlier systems too-the potentialities of the 405 -line system were fully realized.
Another large-screen demonstration, again on 405 lines, featured the Eidophor projector. Baumann, of the Instut für Technische Physik der ETH, Switzerland, described how a beam of electrons, modulated by the video signal, builds up a picture upon a rastar scanned out on a uniform oil film on a mirror, so causing the oil film to be distorted, by electrostatic effects, in proportion to the modulation. An interference-type optical system, using a Xenon lamp, projects the picture, although in this case the light is reflected from the mirror through "rippled" film. The oil is chosen to have a charge-decay time similar to one field of the picture, and, as it is damaged by the bombardment of the beam it is constantly changed by rotation of the mirror and an oil-circulation system. A recent improvement is the use of focus-modulation of the electron beam by an electrostatic lens: when black is required, the beam is out of focus and has little effect on the oil. However, for high brightness the beam is focused sharply and so exerts a greater effect on the oil film.

Once again, the resolution made full use of the 405line system and it must be recorded that, after demonstrations of both the Eidophor and Marconi projectors, questioners asked whether the standard used was 625 lines. Field interlace was excellent in both cases.
Finally, to conclude this short report, may we ask why the cobbler should always be the worst shod of men? Lecturers were put at a disadvantage by several failures of slide-projection arrangements (it is greatly to their credit that they were not too put out) and the sound reinforcement system behaved, throughout the two days, in a manner that would have disgraced a village fête, let alone the I.E.E.

## POST OFFICE DISTRIBUTION NETWORK

By W. L. NEWMAN*

ALTHOUGH the B.B.C. and I.T.A. are individually responsible for providing and operating their television stations, the links between studios and transmitters and between stations in each network are provided by the Post Office which is, of course, also responsible for the links between the B.B.C.'s sound broadcasting stations.

The development of the Post Office television network commenced in 1937, the year after the start of the television service, when a cable of the balanced pair type was laid between various points of interest in London. This ran between Westminster Abbey, Buckingham Palace, Hyde Park, Broadcasting House and the transmitter at Alexandra Palace and was used to televise the Coronation in that year.

The first inter-city television link was provided in 1949 berween London and Birmingham. This has now grown to a national network covering the whole of Eritain and Northern Ireland, with a connection to Lille in France fo-ming the British Post Office component of the Eurovision chain. It includes 2,500 miles of cable and 2,200 miles of radio circuits used for the distribution of the vision signals. Corresponding networks have been set up for the associated sound and control lines, which are an essential part of the circuits provided for the broadcasting authorities, and these, for the B.B.C. and I.T.A. services combined, account for 10,000 circuit miles. The -sound and control circuits are provided over cable links, even when the vision circuits with which they are associated are carried by a radio system.

The main Post Office intercity circuits are routed between

[^5]
(Crown copyrighl reserved)
The radio and cable links in the U.K. television network. Where a station relies on the reception of a nearby transmitter for its programmes (in Post Office parlance, R.R.B.-radio re-broadcast) no link is shown. The links between the terminal points and the transmitters are a combination of radio and cable.

Network Switching Centres (N.S.Cs), of which there are about 12 covering the country. Of these, London, Birmingham, Manchester and Carlisle are the largest. Between the N.S.C.s are provided a number of vision and sound channels in each direction; these can be interconnected to provide tandem connections between selected points. The vision channels must have a $3-\mathrm{Mc} / \mathrm{s}$ bandwidth to carry a 405 -line signal with minimum distortion and the sound channels $10 \mathrm{kc} / \mathrm{s}$. The inter-city channels are rented by either the B.B.C. or the I.T.A. Radiating from each centre are a number of circuits to and from the local studios and broadcast transmitting stations, each link being rented by the appropriate authority. In addition, the B.B.C. and the programme contractors, who provide the I.T.A. programme material, usually have a Programme Switching Centre (P.S.C.) in the main towns. The function of these centres is broadly similar to that of the N.S.C., but deals with a small group of studio connections, and a circuit to the local broadcast transmitter via the N.S.C.

## Circuit Routing

The day to day work performed by the Post Office, in setting up and controlling the circuits, varies according to the authority concerned. The B.B.C. television service is essentially a national one, with only a small number of locally produced and transmitted programme items. Therefore, the circuits rented by the B.B.C. are connected from the N.S.C. through to the Corporation's premises; the B.B.C. then retains control of them, unless a fault develops, until the completion of the day's programme. Any inter-connection of circuits is performed by the B.B.C.'s own engineers, as required.

The I.T.A. service is based on regional units, each region being served by a programme contractor, the I.T.A. being responsible only for renting the intercity and broadcast transmitter circuits. The individual programme contractors rent from the Post Office the circuits to link their studios and P.S.Cs. The I.T.A. programme in any one region is provided by the local programme contractor who is free to sell or buy programme material from any other contractor in the country. To inter-connect the various regions, the I.T.A.-rented circuits are used, and it is the responsibility of the Post Office to link these circuits as required. This is done on a time basis, so that the appropriate programme material is available at the right time and place.
The Post Office also provides circuits for organizations who perform specialized work for the various companies, one such unit provides a video-tape recording service.
The Post Office may be asked to establish a circuit between a studio and a recording unit so that a particular programme may be recorded for use later or for sending abroad for use by other television services. Other companies provide studio facilities which can be hired for use by advertising agents or for closed-circuit television demonstrations, here again the Post Office provides the links.
Another type of circuit requirement is the provision of channels to viewing rooms. These circuits are provided on normal telephone plant using ordinary subscriber's pairs, outside broadcast amplifiers being used to provide the necessary equalization
and gain for a bandwidth of $2.5 \mathrm{Mc} / \mathrm{s}$. These rooms are usually owned $u$ rented by programme contractors whose operational region is not London. They are used to show either broadcast or closedcircuit programmes from their own studios and the Post Office has to provide a circuit from, say, Manchester, to the appropriate viewing room in London.

The main distribution, as explained earlier, is by means of cable and radio system. In general coaxial cables are used for the land line portion of the vision circuits, but there are a few balanced pair cables still in use. The vision signals can readily be carried over coaxial cables at video frequency up to a distance of 15 miles and development work indicates that this will soon be extended.
The main connections between the studios, broadcast transmitter and the programme switching centres in one area radiate from the network switching centre using unbalanced transmission on coaxial cables. The size of the cable may be 0.163 in , 0.375 in or 0.975 in in diameter, depending on the installation, the equalization and gain in the amplifiers being adjusted according to the type used. One of the chief disadvantages of this type of transmission was the induced $50 \mathrm{c} / \mathrm{s}$ signal from mains cables, but this difficulty has been overcome by the use of large coaxial chokes at each amplifier point which effectively suppress the longitudinal current in the cable sheath. The signals carried on the coaxial cables are amplified at intervals, usually in telephone exchange buildings, if one is conveniently placed, or in repeater stations erected for this purpose.
For longer distance transmission, because of interference to signals below $10 \mathrm{kc} / \mathrm{s}$ on coaxial cables, the vision signals are shifted to a higher frequency band using a vestigal sideband transmission system. By a modulation process the video signal is transformed to another range, the limits of which vary with the system employed. Various line frequency standards are employed, the most common being $0.5-4.0 \mathrm{Mc} / \mathrm{s}$ with a $1,056 \mathrm{Mc} / \mathrm{s}$ carrier frequency, whilst the first system between London and Birmingham has a $3-7 \mathrm{Mc} / \mathrm{s}$ band and a $6.12 \mathrm{Mc} / \mathrm{s}$ carrier frequency. At the receiving end of a longdistance system a demodulation process restores the video signal. All main line transmission systems have duplicated equipment to guard against failure.
The radio systems vary because of rapid developments in this field, but the main equipments are built to operate in the 2000 and $4000 \mathrm{Mc} / \mathrm{s}$ bands, with an intermediate frequency of $70 \mathrm{Mc} / \mathrm{s}$. These systems are frequency modulated using equipment designed in the Post Office Research Station. The original London-Birmingham radio system which was recently replaced by more modern equipment, used carrier frequencies in the $900 \mathrm{Mc} / \mathrm{s}$ band and an intermediate frequency of $34 \mathrm{Mc} / \mathrm{s}$.
In the case of the radio equipment some of the earlier systems have complete standby transmitters and receivers, the later systems, however, usually have a complete reserve available from end to end known as a protection channel. This is a separate link working at a different carrier frequency and can be taken into service, either automatically or by manual operation, in the cvent of a failure.

The London Television Network Switching Centre is situated in part of the Museum Exchange building and comprises a repeater station where the cable systems terminate, a radio station on the roof,


Permanent cable links from the London Network Switching Centre to the television transmitters, studios and centres of entertainment in the London area.
a sound amplifier room, and a control room. The repeater station, besides housing the amplifier and translation equipment, has extensive supervisory apparatus to assist the maintenance engineers in their work. The main cable systems are provided with amplifiers at distances of either three, six or 12 miles, mainly in unattended stations. Any abnormal condition in these stations is signalled over a supervisory circuit to the Centre and remedial action is started. Where there are reserve amplifiers in distant stations they are automatically switched into circuit to cure a fault, or they can be operated remotely from the Centre over other supervisory pairs. Power faults in remote stations are also signalled and standby diesel-powered generators automatically start and supply power if the local mains supply fails. Extensive supervisory equipment is also provided on all the radio systems so that faults can be quickly detected and corrected. On some systems the switching of the protection channel is controlled by detecting the $10.125 \mathrm{kc} / \mathrm{s}$ synchronising pulses at both input and output of the system and using these signals to control the switching circuit.

## London Focal Point

The Control Room in the London centre is the terminal point of the incoming and outgoing vision links, which now number over 100. In addition there are over 250 sound and 300 control lines. This room is the focal point of the work in the station and the point from which the cable and radio links are operated. Upon the engineers here and their colleagues in other stations, depend the television ser-
vices for the country. Two test consoles in the room are equipped with the apparatus to carry out the routine testing of the circuits.
The engineers have a group of 15 picture monitors in front of them which allow them to keep a constant watch on the signals passing through the station. In addition they have monitors and c.r.o.s built into the consoles so that dctailed examination of a signal can be undertaken. From the video distribution rack behind them, the engineers can instantly connect any circuit to their test equipment.

On a number of vision circuits, remotely operated vision test signal generators are provided, so that a maintenance engineer is not required to visit distant stations for routine work. One such circuit provided with this type of equipment is the circuit from London Airport which is rented by the B.B.C. When this is required for service, at any hour of the day or night, operation of the appropriate switches will connect a waveform generator in Queens Building to the outgoing circuit. On completion of the liningup procedure, the circuit will be connected direct to the camera by restoration of the switches. Between London Airport and the London N.S.C. there are four intermediate amplifier points, each with reserve equipment. These reserve amplifiers can also be switched into circuit from the control room. Tests are applied to all sound and vision circuits at regular intervals, vision circuits being tested daily.

In a side room off the control area is the Automatic Network Distribution Equipment (A.N.D.E.). This is provided to switch the I.T.A.-rented inter-city links to the various programme contractors who wish to send programmes to, or receive programmes from,
the I.T.A. network. Incidentally, "networking" has resulted in the growth of switching operations from 136 in May, 1956, to 4,540 in April, 1961. The equipment is divided into sound and vision components and can cater for changes of routing from any of 15 sources to any one or all of 15 destinations, in any order. Three changes of routing can be set up at the beginning of each day and at any time subsequently two changes are always available in advance. The times at which changes can be made can be set to take place at intervals of ten seconds if necessary. The electronic clock mechanism, which is driven from the $50 \mathrm{c} / \mathrm{s}$ mains supply, is automatically synchronized by TIM every ten seconds. To ensure complete reliability, two clocks are provided and the indications of these are continuously compared electronically, any deviation being detected and the faulty clock switched out of service until the fault is corrected, during which time manual switching operations are carried out. Instructions for the operation of the equipment are received from the I.T.A. on a daily schedule; this is amended as required during the day. A news item of major importance can mean the complete rearrangement of an evening's switching schedule; fortunately this is a rare occurrence. Switching operations have to be carried out in a period of three seconds and are usually completed ten seconds before a minute and frequently are required to provide networking conditions for the start of a new programme concurrent with the end of the previous item.
The Automatic Network Distribution Equipment incorporates two vision picture monitors, one of which can be switched by the controlling engineer to any of the signal sources and the other to the destination signals. In addition, a high-quality music amplifier and loudspeaker allows the various sound signals to be monitored. The engineer is able to check both sound and vision before and after a switch operation to ensure that it has been successfully completed.
On coaxial cable systems, to reduce the risk of breakdowns to equipment in remote repeater stations
due to mains supply failures, it is only at what are known as power-feeding repeater stations that power is taken from the local supply. These repeaters are usually in towns and from these power is fed at high voltage over the coaxial cable itself to the other repeater stations. At the dependent stations the supply is transformed to a voltage suitable for the local amplifying equipment. At the power-feeding stations, diesel-powered generators are installed and these are automatically switched into scrvice in a period of ten seconds in the event of a mains failure.

Coaxial cables are liable to damage due to road works and by working parties of other public utilities. To give early warning of damage which might interrupt a service, all cables are continually monitored to check their insulation. In addition, on the latest installations, the cables are kept under pressure from a gas reservoir and any damage causes an alarm indication to be given. Also, the escape of the gas limits the amount of damage which can be caused by water, if present. Everything is done to reduce the risk of accidental damage but emergency repair teams are available in all parts of the country to go to the site of any damage. In London the external repair team can keep in touch with the Centre by means of mobile radio equipment in their vans. The fixed station on this system is at Harrow Weald, which is the first intermediate amplifier point on the London-Birmingham radio link system; this is controlled over a land line from the London Network Switching Centre.
Besides the permanent television distribution network, the Post Office also provides a large number of vision and sound circuits on a temporary basis for use from outside broadcast sites. Some of these sites are in regular use and permanent cable facilities are provided, the amplifying equipment being provided only for the duration of the broadcast, however. Other broadcasts need the use of temporary circuits established on ordinary underground telephone wires. Such circuits can be provided using up to three miles of cable, with variable equalization amplifiers at mile intervals. Some O.B.s use a combination of various types of transmision equipment, perhaps starting from a site using a normal telephone pair to the nearest hilltop, where the signal is injected into a portable microwave radio link, which carries the signal to a receiver positioned near a repeater station on a permanent coaxial cable system, where the signal, in turn, is fed into the main distribution network.
In addition to providing vision circuits for the present standards, a number have been installed to carry 625 -line signals. Some of these circuits are used for recording purposes, the tapes being intended for use abroad, others are for closed-circuit work.

Combined vision and sound control position at the London Network Switching Centre which is housed in the Museum Telephone Exchange off Tottenhom Court Road.

## NEW ELECTRONIC EQUIPMENT AND ACCESSORIES

## High-impedance Voltmeter

THE Electrosensor voltmeter developed by the American firm of Halex is, in essence, a highly sensitive, unity gain amplifier. The electrometer-valve input stage is followed by three unity gain cathode-followers, the earth of each forming the screen for the succeeding stage. Each screen is connected to one of three guard rings on the coaxial input socket, and as each has impressed on it a voltage very nearly equal to that at the input, the impedance is very considerably increased. This same technique also serves to decrease the input capacitance, and the overall effect is a very highimpedance, short time-constant input.
The output of the amplificr is at low-impedanceless than $0.2 \Omega$-and is used to drive a meter, switchable to read from 1 V to 250 V d.c. full-scale, or 500 V p.p. for very low frequency a.c. The input impedance is $10^{16} \Omega$, and the input capacitance less than 0.01 pf . Drift after warm-up is 2 mV per hour
Accessories are available to enable the instrument to function as an ohmmeter, a static charge indicator, a current regulator, etc.
The instrument is marketed in the United Kingdom by Scientific Furnishings, Ltd., West Hampnett Road, Chichester, Sussex.

## Motor-driven TV Tuner

RECENTLY introduced by the Plessey Company, this television tuner is driven by a mains-powered induction motor to avoid the interference problems that would be raised by a commutator type. An automatic clutch disconnects the drive when the motor is switched off,


Plessey motor-driven turret tuner.
allowing the indexing mechanism in the tuner to exert full control. Remote push buttons or a rotary switch can be used for channel selection and the tuner has the valuable advantage that, as it does not require fine tuning, it can be placed anywhere in the receiver.

The tuner uses frame-grid valves and Band II v.h.f., as well as TV, channels can be incorporated. A complete rotation takes only seven seconds.

The Plessey Co., Ltd., Ilford, Essex.

## Closed-circuit Television Camera

THE "Nashton" Mark I industrial closed-circuit tclevision camera provides a signal which can be fed directly into a standard 405 - or 625 -line domestic receiver (preferably without flywheel sync) covering Band I. Costing £275, this camera is only $13 \frac{1}{2} \times 7 \frac{1}{4} \times 7 \frac{1}{4}$ in and is completely self-contained. A Vidicon-type tube

Nashton Mk 1 closedcircuit television camera.

is used and the standard lens fitted to the four-position turret has an aperture of $f / 1.9$, the lin focal length giving a horizontal field of view of $28^{\circ}$ (at loft this corresponds to an area 5 ft by 3 ft 9 in ). As is common practice, the time bases used are not locked togetherthe resulting pictures are thus not interlaced. Resolution is about 300 lines overall.
Nash and Thompson, Ltd., Hook Rise, Tolworth, Surrey.

## Safe Grease Solvent

CHLOROTHENE NU, marketed by PenetoneParipan, is a form of methyl chloroform and has valuable advantages as a solvent for oils, greases and waxes.

The solvent evaporates quickly and is claimed to leave no residuc liable to cause tracking. In addition it is far less toxic than the ubiquitous "carbon tet." Comparable maximum allowable atmospheric concentrations are 25 parts in $10^{n}$ for carbon tetrachloride, and 500 parts in $10^{6}$ for methyl chloroform. It is noninflammable.
Penetone-Paripan, Ltd., Egham, Surrey.

## Oscilloscope Cameras

AVELEY Electric are to market two oscilloscope cameras manufactured by Du Mont. The type 299, which accepts 120 roll- or cut-film, is for general recording from a 5 in tube, while the type 302 incorporates a Polaroid-Land back, and will deliver a finished print


Du Mont type 302 cscilloscope camera. A pointer shows the position of the sliding back in relation to the lens.
on transparent or opaque film within one minute. Both cameras have a sliding back which facilitates the recording of more than one trace on each frame. Either an $\mathrm{f} / 2.8$ three-element Wollensak Du Mont or an $\mathrm{f} / 1.9$ sixelement lens is available, both of 75 mm focal length. Adaptors enable the camera to be used with most 5 in oscilloscope tubes. Aveley Electric, Ltd., Ayron Road, Aveley Industrial Estate, South Ockendon, Essex

## Saw for Cutting Sheet Material

THE ordinary hacksaw is inconvenient for the hand cutting of sheet metal or insulating material because the frame gets in the way-this can lead either to the use of blades without a frame or a wood saw, neither of which is the best of tools for the job. Stead's


## "Steadfast" sheet saw.

"Steadfast Sheet Saw" employs replaceable 12 in-long triangular blades ( 14 or 24 teeth/in), supported by a "spine"-the edge bearing the teeth is thus presented at the correct angle for cutting without the necessity for the spine to pass through the material being cut. The saw can be used for cutting unlimited length of flat or corrugated sheet as well as being suitable for most of the jobs for which a hacksaw is normally used.
J. Stead and Co., Ltd., Manor Works, Cricket Inn Road, Sheffield, 2.

## Pulse Generator

WITH rise- and fall-times of 15 and 20 millimicroseconds respectively, the Wayne-Kerr Short Duration Pulse Generator P131 has many applications in computer development and pulse work generally. The pulse recurrence frequency may be from 50 p.p.s. to 50 k.p.p.s., while the pulse widths available are from $50-1,000$ millimicroseconds. The output amplitude is variable in 0.5 dB steps between 0.1 V and 10 V at an impedance of 75 ohms, while a high level output of up to 40 V is available at a source impedance which varies between 50 and 100 ohms. Facilities for pre-pulse, delay and external triggering are provided. Details are obtainable from Wayne-Kerr Laboratories, 44, Coombe Road, New Malden, Surrey.

## High-speed Printer

A PRINT-OUT unit capable of operation at a speed of over 10 lines/second has been developed by Solartron. Fourteen columns are printed by continuouslyrotating print wheels, which are struck at the required times by hammers. No reset mechanism is employed; as the next available character moves to the correct position on any print wheel, printing begins, no waiting being necessary. The equipment is in two sectionsthe printing machine proper and the electronic rack in a separate case. Inputs may be either decimal or binary at a voltage of -6 V to -12 V and the print command signal required is a $6 \mathrm{~V}, 1 \mathrm{msec}$, negative-going pulse. The equipment is produced by the Solartron Electronic Group, Ltd., Victoria Road, Farnborough, Hants.

## Wide Band Oscilloscope

THE Cossor 1076 oscilloscope continues the trend towards a high quality basic instrument with plug-in extension facilities. An extremely wide frequency response (d.c. $-60 \mathrm{Mc} / \mathrm{s}$ at -3 dB ) is coupled with a sensitivity of $50 \mathrm{mV} / \mathrm{cm}$, which it is possible to increase to $5 \mathrm{mV} / \mathrm{cm}$ while still retaining a bandwidth of d.c. $-40 \mathrm{Mc} / \mathrm{s}$. Voltage and time are calibrated within $\pm 2 \%$ and $\pm 3 \%$ respectively, and a useful feature is the provision for Z -modulation at frequencies of $50 \mathrm{Mc} / \mathrm{s}$ and $500 \mathrm{Mc} / \mathrm{s}$.

The 5 in -diameter tube is operated at 10 kV , the useful area being $6 \mathrm{~cm} \times 10 \mathrm{~cm}$. All the usual facilities are pro-vided-illuminated graticule, spot-position indicators, etc., and outputs provided are the amplified Y signal, and the gate and sawtooth waveforms of the timebase. Signal delay, of 150 msec , is by a specially developed distributed-constant delay cable.

Extension Y facilities available include a high-gain amplifier, which affords a sensitivity of $1 \mathrm{mV} / \mathrm{cm}$ from d.c. to $1 \mathrm{Mc} / \mathrm{s}$, and which has two input sockets. The inputs may be viewed separately, or combined differentially, in which case the common-mode rejection ratio is of the order of 80 dB up to $50 \mathrm{kc} / \mathrm{s}$. A dual channel unit is also available which provides for the viewing of two inputs separately, algebraically added, or simultaneously by chopping or beam-switching on alternate sweeps.

A comprehensive range of triggering and delay facilities is provided by three X units. In addition to the normal triggering and windowing modes at a minimum signal level of 2 mm deflection, or 0.2 V external trigger, two calibrated units provide for sweep delays of from $2 \mu \mathrm{sec} / \mathrm{cm}$ to $1 \mathrm{sec} / \mathrm{cm}$, with the delaying sweep brightened over the portion to be expanded.

The instrument is manufactured by Cossor Instruments, Ltd., Cossor House, Highbury Grove, London, N.5. The oscilloscope with the wide-band preamplifier and general-purpose trigger unit costs $£ 650$.


The Cossor 1076 wide band oscilloscope, with its range of plug-in units, which extend the $X$ and $Y$ performance.

# Transformer-Ratio-Arm Bridges 

USE IN THREE-TERMINAL IMPEDANCE MEASUREMENTS

By J. F. GOLDING*

BRIDGES of the transformer-ratio-arm type are not by any means new. Indeed, the system was the subject of a patent by Blumlein as long ago as 1928. The author first encountered a bridge of this type-an armed-forces instrument designed by E.M.I.-during the early years of the war. It was then known as the in situ bridge because its three-terminal-measuring facility permitted measurement of a circuit-component's value without disconnecting it.

The transformer-ratio-arm bridge has a number of other important attributes; but, as the threeterminal facility is probably the most impressive, let us consider the derivation of the bridge from this point of view.

Three-Terminal Measurements.-The requirement for three-terminal measurement arises when the impedance to be measured is in parallel with some unwanted impedance. A typical example is the stray shunt capacitance between test leads or terminals that often makes measurement of low capacitance somewhat uncertain. The effect of this unwanted shunt impedance can be eliminated if it can be divided into two elements in series, with their junction connected to a neutral point in the bridge.

The rudiments of this kind of three-terminal measurements can be illustrated by examination of the circuit arrangement shown in Fig. 1. $G$ is a zero-impedance voltage source generating an e.m.f., $E$, which causes a current I to flow in the unknown impedance, $\mathrm{Z}_{u}$. This current is monitored by a zero-impedance meter $M$. Thus the impedance of the unknown is obviously given by $\mathrm{E} / \mathrm{I}$.

But the dotted lines show, in parallel with $Z_{u}$, the series combination of two unwanted impedances, $Z_{1}$ and $Z_{2}$. The effect of this shunt impedance is, however, obviated by connecting the junction of $Z_{1}$ and $Z_{2}$ to the neutral linc. $Z_{1}$ is then connected directly across the source, and $Z_{2}$ is across the current


Fig. I. Basic three-terminal-measurement system.
monitor. It is manifest that, since the source has zero impedance, the p.d. across $Z_{1}$ must be equal to $E$; and, as no p.d. can be developed across the zero-impedance meter, there is no current flowing in $Z_{2}$. So neither of these unwanted impedances can effect the accuracy of the measurement.

That this idealized arrangement could not be used in practice is too obvious to justify further comment; but the principle demonstrated can be applied to practical bridge systems.

The arrangement of a conventional impedance bridge with resistive ratio arms is shown in Fig. 2(a). This form is very well known, the conditions for balance being $\mathbf{Z}_{u} / \mathbf{Z}_{s}=\mathbf{R}_{1} / \mathbf{R}_{2}$, where $\mathbf{Z}_{u}$ is the unknown, $\mathbf{Z}_{s}$ is the standard impedance, and $\mathbf{R}_{1}$ and $\mathrm{R}_{2}$ are the ratio arms.

Applying Thevenin's theorem, the equivalent circuit shown in Fig. 2(b) can be derived. Here the resistive ratio arms are replaced by two e.m.f. generators producing voltages, $\mathrm{E}_{1}$ and $\mathrm{E}_{2}$, equal to those appearing across the respective ratio arms at balance. The equivalent resistance of the ratio arms is represented by $\mathbf{R}_{r}$ between the detector and the junction of the two generators, the value of $\mathrm{R}_{r}$ being that of the two ratio-arm resistors in parallel.

It is conceivable that the values of the ratio-arm resistors could be made so small that $\mathrm{R}_{r}$ became negligible compared with the other impedances in the circuit. The equivalent circuit in Fig. 2(b) could then be redrawn as shown in Fig. 2(c).

In this last diagram the series combination of unwanted stray impedances, $\mathrm{Z}_{1}$ and $\mathrm{Z}_{2}$, is also shown in dotted lines, their junction being taken to the point in the circuit where the detector joins the ratio arms; i.e., point C in Figs. 2(a) and 2(b).

Part of the circuit in Fig. 2(c) is drawn in heavy line; and this part is recognizable as being very similar to the circuit in Fig. 1. The essential difference lies in the fact that, in Fig. 2(c), the current in the unknown impedance is not monitored directly; instead it is balanced by the current in the standard impedance $Z_{s}$. But the same arguments regarding the cancellation of the unwanted stray impedance apply to both circuits. Although, in practice, the detector in Fig. 2(c) would not have zero impedance, this would not affect the accuracy of three-terminal measurements because no p.d. is developed across the detector when the bridge is balanced, and thus no current flows in $\mathrm{Z}_{2}$.

However, reduction of the ratio-arm resistance to the point where $\mathrm{R}_{r}$ is negligible is not practical in a conventional bridge; for, to obtain a workable

[^6]
(a)
(b)

(c)

Fig. 2. (a) Conventional Impedance Bridge. (b) Equivalent circuit of conventional bridge of Fig. 2(a). (c) Equivalent circuit of conventional bridge when $R_{r}$ of Fig. 2(b) is negligible.
sensitivity with such an arrangement, the current in the ratio arms would have to be very high indeed. And, if $\mathbf{R}_{r}$ has an appreciable value, the stray impedance $\mathrm{Z}_{1}$ appears across the generator $\mathrm{E}_{1}$ and $\mathrm{R}_{r}$ in series. The current flowing in the unknown and the detector due to $\mathrm{E}_{1}$ is then reduced by the loading of $Z_{1}$, thus affecting the accuracy of the bridge.

It becomes evident, therefore, that in the practical bridge the impedance of the detector is of secondary importance from the point of view of threeterminal measurements, whereas the impedance of the ratio arms is very important. And the problem resolves into one of producing two voltages, having a constant ratio, from very-low-impedance sources.

This is easily achieved by replacing the resistive ratio arms by a pair of secondary windings of a voltage transformer as shown in Fig. 3. Such a bridge closely approaches the ideal of Fig. 2(c). For, providing the leakage inductance and losses in the voltage transformer are small, the loading of an unwanted impedance, $Z_{1}$, across one ratio-arm winding is distributed proportionally between the windings, and the voltage ratio remains constant.

## Circuit Arrangements

Tapped Voltage Transformer.-In the basic circuit of Fig. 3, the standard impedance, $Z_{s}$, must be varied to achieve balance. Indeed, the impedanc


Fig. 3. Basic circuit of transformer-ratio-arm bridge.
range of the bridge is determined by the range over which the standard can be varied.

A variable-impedance standard is costly, complex, and susceptible to errors. Many conventional bridges are therefore built with fixed standards and variable ratio arms; and this system is ideally suited to the transformer-ratio-arm bridge. Fig. 4(a) shows a simple bridge in which a fixed standard is used and the voltage ratio is varied by the use of a tapped winding on the transformer.

The number of turns in the winding feeding the unknown is $N_{1}$, and the winding feeding the standard impedance is tapped at $\mathrm{N}_{2}$ turns, so that the conditions for balance are:-

$$
\begin{equation*}
\frac{\mathrm{N}_{1}}{\mathrm{~N}_{2}}=\frac{\mathrm{Z}_{u}}{\mathrm{Z}_{s}} \quad \text { or } \quad \mathrm{Z}_{u}=\frac{\mathrm{N}_{1} \mathrm{Z}_{s}}{\mathrm{~N}_{2}} . \tag{1}
\end{equation*}
$$

Thus, the effect of switching the standard impedance through 10 tappings, spaced equally over the winding as shown in Fig. 4(a), is equivalent to changing its value over a range of 10 to 1 in equal steps.

By using a number of standard impedances having decade relationships to each other a multipledecade bridge can be made. Such an arrangement is shown in Fig. 4(b). This diagram shows three standards $Z_{s 1}, Z_{s 2}$, and $Z_{s 3}$, whose values are such that $Z_{s 1}=10 Z_{s 2}=100 Z_{83}$. The effective admittances$\mathrm{N}_{2} / \mathrm{N}_{1} \mathrm{Z}_{s}$-of these standards can be added together.

To simplify the expression for balance conditions of such a bridge, let the ratio $\mathrm{N}_{1} \mathrm{~N}_{2}$ for $\mathrm{Z}_{s 1}$ be written $n_{1}$, that for $Z_{s 2}$ be $n_{2}$ and that for $Z_{s 3}$ be $n_{3}$. The expression then becomes:

$$
\begin{equation*}
\frac{1}{\mathrm{Z}_{u}}=\frac{1}{n_{1} \mathrm{Z}_{s 1}}+\frac{1}{n_{2} Z_{s 2}}+\frac{1}{n_{3} Z_{s 3}} . \tag{2}
\end{equation*}
$$

Tapped Current Transformer.-If the bridge is intended to cover a wide range of impedances or if more than three significant figures are required, a practical difficulty arises due to the wide divergence of standard-impedance values. This is overcome by feeding the detector via a current transformer with a tapped primary winding as shown in Fig. 4(c).

Here the unknown is connected at $\mathrm{N}_{3}$ turns, and the standard is connected to a tap at $\mathrm{N}_{4}$ turns. The voltage across the secondary winding of the current transformer is proportional to the difference between the flux due to the current in $Z_{u}$ and that due to the current in $\mathrm{Z}_{s}$. This can be expressed as:-

$$
\frac{\mathbf{N}_{1} \mathbf{N}_{3}}{\mathrm{Z}_{u}}-\frac{\mathbf{N}_{2} \mathrm{~N}_{4}}{\mathrm{Z}_{s}}
$$



Fig. 4. (a) Use of a tapped-ratio-arm winding instead of a variable standard. (b) Use of three standards to give a three-decade read-out. (c) Use of a tapped current transformer T2 to increase the effective value of the standard $Z_{3}$. (d) Application of Fig. 4(c) to give a four-decade read-out.

Equating this to zero, the balance-condition expression becomes:-

$$
\begin{equation*}
Z_{u}=\frac{\mathrm{N}_{1} Z_{s} \mathrm{~N}_{3}}{\mathrm{~N}_{2} \mathrm{~N}_{4}} \tag{3}
\end{equation*}
$$

Comparing expression (3) with expression (1), it is evident that the effective value of $Z_{s}$, has been changed by the ratio $\mathrm{N}_{3} / \mathrm{N}_{4}$.

It follows, then, that a number of tappings on the current-transformer primary winding can be used in order to provide decade-impedance steps. Fig. 4 (d) shows the application of this reasoning. The primary winding of T 2 has $\mathrm{N}_{3}$ turns, and it is tapped at $\mathrm{N}_{4}$ turns, $\mathrm{N}_{5}$ turns, and $\mathrm{N}_{6}$ turns, where $\mathrm{N}_{4}=$ $\mathrm{N}_{3} / 10, \mathrm{~N}_{5}=\mathrm{N}_{3} / 100$ and $\mathrm{N}_{6}=\mathrm{N}_{3} / 1000$.

Four standards are used having an equal impedance value, $Z_{s}$. But, due to the action of $T 2$, their effective impedances are in decade relation. The balancecondition expression is thus a composite of expressions (2) and (3) as follows:

$$
\begin{equation*}
\frac{1}{Z_{u}}=\frac{1}{n_{1} Z_{s}}+\frac{\mathrm{N}_{4}}{\mathrm{~N}_{3} n_{2} Z_{s}}+\frac{\mathrm{N}_{5}}{\mathrm{~N}_{3} n_{3} Z_{s}}+\frac{\mathrm{N}_{6}}{\mathrm{~N}_{3} n_{4} Z_{s}} \tag{4}
\end{equation*}
$$

It has already bcen stated, however, that the turns at $\mathrm{N}_{3}, \mathrm{~N}_{4}, \mathrm{~N}_{5}$ and $\mathrm{N}_{6}$ are in decade relationship. This being so, the general expression (4) can be simplified to:-

$$
\begin{equation*}
\frac{1}{\mathrm{Z}_{u}}=\frac{1}{\mathrm{Z}_{s}}\left(\frac{1}{n_{1}}+\frac{1}{10 n_{2}}+\frac{1}{100 n_{3}}+\frac{1}{1000 n_{4}}\right) \ldots \tag{5}
\end{equation*}
$$

Complex Impedances.-Any complex impedance measured at a single frequency can be resolved into an equivalent series or parallel network; and it has already been shown that the transformer-ratio-arm bridge indicates the sum of the effective admittances of its standards. So it is logical to regard the impedance being measured as a reactance in parallel with


The Marconi Low-Capacitance Bridge Type TF 1342 is suitable for the measurement of capacitance down to $0.002 \mu \mu F$. It is a transformer-ratio-arm bridge of the form shown in Fig. 4(d).
a resistance. These two parts are then balanced against reactive and resistive standards respectively.

It appears then that the transformer-ratio-arm principle is particularly applicable to multi-decade capacitance bridges. For example, expression (5), rewritten for the measurement of pure capacitance, simplifies to:-

$$
\mathrm{C}_{u}=\mathrm{C}_{s}\left(\frac{1}{n_{1}}+\frac{1}{10 n_{2}}+\frac{1}{100 n_{3}}+\frac{1}{1000 n_{4}}\right)
$$ where $C_{u}$ is the unknown capacitance and $C_{s}$ is the value of the capacitance standards.

Multi-decade resistance or inductance measurement, however, is not nearly so straightforward. Indeed, where the required accuracy is such that a multi-decade system is necessary, the bridge is usually calibrated in terms of conductance and susceptance rather than resistance and reactance.

Continuously-Variable Controls. - Where a high degree of accuracy is not required, a single contin-uously-variable balance control is sometimes preferable to a series of decade switches. This applies particularly to resistance and inductance measurement; for a variable control with direct calibration is likely to be more convenient than switched decade controls with reciprocal calibration. A continuouslyvariable control is also usually desirable to indicate the last significant figure on a bridge with decade controls.

There are two ways of providing continuous variation. The more obvious is by the use of a variable standard; i.e., a variable capacitor or resistor. The other method, which is more convenient when a high effective impedance is required, is shown in Fig. 5.

Here a variable potentiometer is connected across the ratio-arm winding, and the standard impedance is connected to its slider. The action of varying the potentiometer is then exactly analagous to that of switching through a series of tappings on the transformer winding. And, providing the resistance of the potentiometer is small, the behaviour of the ratio arm remains that of a low-impedance e.m.f. generator.

Very high effective impedances can be produced by this second method since the voltage at the slider can be reduced almost to zero. In fact this method gives an effective impedance many times greater than could reasonably be attained with a variable impedance connected to the end of the winding.


Fig. 5. Continuously-variable standard arm.

Correction of Standard Impedances.—Although high-quality components are naturally used for the standard impedances, some phase-angle errors in the standards are inevitable; a standard capacitor is bound to have some loss, and the standard resistor may possibly have some stray reactance.

Such errors can easily be compensated for by the deliberate introduction of the right amount of loss or stray reactance into the opposite arm of the bridge as shown in Fig. 6(a). In this diagram $\mathrm{C}_{s}$ is the standard having an equivalent shunt loss of $\mathrm{R}_{\mathrm{L}}$. Resistor $\mathrm{R}_{c}$ balances $\mathrm{R}_{\mathrm{L}}$ so that the resistive component of $Z_{u}$ is balanced againts the standard resistance $\mathrm{R}_{s}$. The value of $\mathrm{R}_{r}$ is obviously given by:-

$$
\mathrm{R}_{o}=\frac{\mathbf{N}_{1} \mathrm{R}_{\mathrm{L}}}{\mathrm{~N}_{\mathrm{z}}}
$$



Fig. 6. (a) Use of a compensating resistance $R_{e}$ to balance the loss, $R_{\mathrm{L}}$, in the capacitive standard. (b) Reversed arrangement to Fig. $6(a)$ keeping a constant ratio between $R_{L}$ and $R_{c}$.

But, referring to a practical arrangement of the bridge as shown in Fig. 4, an obvious difficulty becomes evident. For balance is achieved by varying the ratio $N_{1} / N_{2}$ and this implies a different value of $\mathbf{R}_{c}$ for each setting of the balance control.

The difficulty can be overcome by rearranging the bridge as shown in Fig. 6(b). Here the transformers have been interchanged; the ratio arms are in the current transformer and a single voltage-transformer winding feeds both the standard and the unknown impedance. With this arrangement, the ratio $\mathrm{N}_{1} / \mathrm{N}_{2}$ remains constant for all settings of the balance-control switch; so a
(Continued on page 333)
single value of $\mathrm{R}_{c}$ can be selected to offset the loss in the standard capacitor.

Fig. 6(b) also shows a preset-variable-adjustment arrangement, obviating the need for individual selection of $\mathrm{R}_{c}$. This uses the principle illustrated in Fig. 5. The potentiomcter is adjusted during the initial calibration of the instrument to give correct phase balance using a capacitor with known shunt loss connected to the unknown terminals.

This inverted arrangement also simplifies the design of a multi-decade bridge for measurement of both inductance and capacitance. Capacitive standards are almost invariably used, and, for inductance measurement, they are switched into the same arm of the bridge as the unknown.

Inductance is then indicated in terms of " negative capacitance", which may be converted by the formula:-

$$
\mathrm{L}_{u}=\frac{1}{\omega^{2} \mathrm{C}_{i}}
$$



Fig. 7. Transformer-ratio-arm bridge for measuring capacitance, inductance and resistance.
( $\mathrm{Z}_{2}$ in Fig. 2) across the detector cannot produce an error.
Regarding the ratio-arm windings of the transformer as voltage generators, leakage inductance and loss can be regarded as an inductance, $L_{l}$, and a resistance, $\mathrm{R}_{t}$, in series. For simplicity let $\mathrm{R}_{t}+\mathrm{j} \omega \mathrm{L}_{t}=\mathrm{Z}_{t}$.
Providing $Z_{t}$ is small compared with $Z_{u}$, the normal condition for balance remains:-

$$
Z_{u}=\frac{\mathrm{E}_{1} \mathrm{Z}_{s}}{\mathrm{E}_{2}}
$$

Any stray impedance, $Z_{1}$, across the winding will, however, drop the voltage applied to the unknown to a value given by $E_{1} Z_{1} /\left(Z_{1}+Z_{t}\right)$. The conditions for balance are then:-

$$
\mathrm{Z}_{u}=\frac{\mathrm{E}_{1}}{\mathrm{E}_{0}} \mathrm{Z}_{s} \frac{\mathrm{Z}_{1}}{\mathrm{Z}_{1}+\mathrm{Z}_{t}}
$$

where $\mathrm{L}_{n}$ is the unknown, $\mathrm{C}_{i}$ is the indicated capacitance and $\omega$ is $2 \pi$ times the measurement frequency.

Bridges employing this system usually operate at such a frequency that $\omega$ is equal to some convenient power of 10 , so that the conversion can casily be made with the aid of a set of reciprocal tables.

An example circuit of a bridge suitable for measuring capacitance, inductance, and resistance is given in Fig. 7. (For simplicity, only two capacitance and two resistance decades are shown.) Such a bridge should be regarded as an admittance bridge and would normally be calibrated in positive and negative capacitance and conductance.

## Quantitative Analysis

Three-Terminal Measurements.-The accuracy of three-terminal measurements with low-value shunt impedances is directly related to loss and leakage inductance in the transformers.

For assessment of the errors caused by shunt impedance, consider first the bridge arrangements shown in Fig. 4.

The simplest practical form of the bridge is that of Fig. 4(b). In this arrangement no current transformer is used, so that any shunt impedance

But $E_{1} Z_{s} / E_{2}$ is the indicated impedance, and calling this $Z_{\text {ind }}$ we have: -

$$
\begin{aligned}
Z_{u} & =Z_{\text {ind }} \frac{Z_{1}}{Z_{1}+Z_{t}} \\
\text { or, } Z_{u} & =Z_{\text {ind }}\left(1+\frac{Z_{t}}{Z_{1}}\right)^{-1}
\end{aligned}
$$

This can be expanded to a binomial series:-

$$
Z_{u}=Z_{\text {ind }}\left(1-Z_{i} Z_{1}+\frac{1}{2!} \cdot \frac{Z_{t}{ }^{2}}{Z_{1}{ }^{2}}-\frac{1}{3!} \cdot \frac{\mathrm{Z}_{t}{ }^{3}}{\mathbf{Z}_{1}{ }^{1}}+\text { etc }\right)
$$

Unless the shunt impedance is so small that it reduces the bridge sensitivity, the second order and subsequent terms can be neglected and the expression written:-

$$
\begin{equation*}
Z_{u}=Z_{i n d}\left(1-\frac{Z_{i}}{Z_{i}}\right) \tag{6}
\end{equation*}
$$

The bridge thus reads high by a percentage error equal to $\left(Z_{t} / Z_{1}\right) \times 100$.
The valuc of $Z_{t}$ of course varies with the number of secondary terms, and this is directly related to the impedance range. However, $Z_{t}$ seldom exceeds 1 part in $10^{4}$ relative to $Z_{u}$, the unknown. Accepting this value of $Z_{i}$, if $Z_{1}$ is as low as a tenth of $Z_{u}$ the error will be only $0.1 \%$.

In practice the output impedance of the transformer is usually much lower relative to the unknown so that even lower value shunt impedances can be tolerated.
Where a current transformer is used as shown in Fig. 4(d), the loss and leakage inductance become significant only in relation to those standards which are tapped down the winding. The error due to shunt impedance across the detector is thus more difficult to evaluate.
However, the worst possible condition occurs when the bridge is operating at the extreme highimpedance end of its range so that the standard impedance tapped lowest on the primary of the current transformer is the only one in use. The conditions in the current transformer are then similar to those in the voltage transformer, and the leakage inductance and loss can also be represented as a series impedance $Z_{i}$.
The current in the transformer winding is thus lowered by the introduction of shunt impedance $Z_{2}$, to a value given by $\mathrm{IZ}_{2} /\left(\mathrm{Z}_{i}+\mathrm{Z}_{2}\right)$. Similar reasoning to that used for assessing the error due to $Z_{t}$ can thus be followed giving:-

$$
\begin{equation*}
Z_{u}=Z_{\text {ind }}\left(1-\frac{Z_{i}}{Z_{v}}\right) . \tag{7}
\end{equation*}
$$

Depending on the position of the tap on the primary windings, $Z_{i}$ may vary from zero to something approaching 1 part in $10^{1}$ relative to $Z_{u}$; and the probable error due to shunt impedance across the detector is likely to be very much less than that due to impedance across the voltage transformer.

Measurement Range.-An important attribute of the transformer-ratio-arm arrangement is its facility for very wide ranges of impedance measurement with a single bridge. Range switching is normally accomplished by switching the unknown impedance to an appropriate tapping on the ratioarm winding, the switch being calibrated as a range multiplier. Such an arrangement is incorporated in the example shown in Fig. 7.

This diagram shows range switching through only one decade; but, assuming that the low-impedance tapping is across 10 turns, the use of three decades is not unreasonable. A 10,000 turn winding would, thus, be required with tappings at 1,000 turns, 100 turns, and 10 turns. Impedance-range multiplication would then be possible from x1 to $\times 1,000$.

Such a multiplier arrangement used in a bridge with four decade controls, as shown in Fig. 4(d), would provide an overall ratio of ten million to one between the lowest and highest impedances that can be measured.

Measurement Accuracy.-The accuracy of any bridge is a function of the accuracies of its standard impedances and its ratio arms. Furthermore, in practice, the attainable bridge accuracy can be assessed in terms of the stability of its components; for, by the use of suitable padding, the value of any component can be effectively corrected to any desired accuracy within the limits of its stability.

In conventional-impedance-ratio-arm-bridges stability of the ratio-arm components is an important factor affecting the accuracy of the bridge. Transformer ratio arms, however, are intrinsically com-
pletely stable; for the turns ratio cannot possibly vary.

The bridge stability is thus the stability of its standards alone. Also, the use of transformer multiplication at the detector as well as at the voltage source permits selection of standard-impedance values for maximal stability regardless of the range of the bridge.

Transformer-ratio-arm bridges having accuracies of the order of $0.1 \%$ are, therefore, simple to construct and are normally modestly priced.

## Application Notes

It is evident from the foregoing that the transformer-ratio-arm system exhibits clear advantages for certain applications. A tendency thus arises to suppose that the system is superior to the conventional bridge arrangement for all applications. This, of course, is not so; and the following comparison notes may be helpful in assessing the suitability of a transformer-ratio-arm bridge for any particular purpose.

Capacitance Measurement.-Unlike resistance and inductance, capacitance is a direct measure of admittance rather than of impedance. As the transformer-ratio-arm bridge is basically more suitable for measurement in terms of admittance, it is automatically admirably suited to capacitance measurement.

Furthermore, it follows that with the transformer-ratio-arm system the loss is more conveniently indicated in terms of equivalent shunt resistance. This is fairly realistic as applied to most capacitance measurements. If the loss is small it is unimportant whether it is regarded as shunt or series loss. In capacitors having an abnormally high power factor, the loss is generally due to poor dielectric so that the representation as equivalent shunt resistance is largely the true one.

The transformer-ratio-arm bridge is particularly suitable for measurement of very-low capacitance values. Such measurements with a conventional bridge are susceptible to errors due to the stray capacitance between connecting leads and between bridge terminals. When short rigid connecting leads are used this stray capacitance is sensibly constant, and errors can be largely eliminated by evaluating the strays and correcting accordingly.

There are many applications, however, where short rigid connections cannot be used. Determination of a capacitor's temperature coefficient is a typical example.
With the component under test housed in a tem-perature-controlled oven comparatively long leads must inevitably be used for connection to the bridge. Such a test is likely to involve a protracted series of measurements; the changes in capacitance may be quite small; and, even though care is taken to prevent physical movement of the leads, the magnitude of possible error due to quite small changes in stray capacitance introduces an element of uncertainty in the final result.
The difficulty can be completely overcome if a transformer-ratio-arm bridge is used. The test leads may then take the form of screened cables with the screens taken to the neutral point of the bridge. The stray shunt capacitance is then divided into two capacitances effectively in series and is
automatically cancelled by the three-terminal measurement.

Inductance Measurement.-Measurement of inductive reactance by means of a transformer-ratioarm bridge is normally made by switching the standard capacitors into the "unknown" arm as shown in Fig. 7.

For bridges having accuracies up to about $1 \%$, the continuously-variable arrangement can be used (see Fig. 5) with a direct inductance calibration on the dial. For higher-order accuracies a multidecade bridge must be used; and the inductance is indicated in terms of "negative capacitance". This is converted to inductance by the formula $\mathrm{L}=1 / \omega^{2} \mathrm{C}$.

Apart from the inconvenience of making this conversion, it is immediately evident that the accuracy of the bridge is very dependent upon the frequency of its exciting voltage. The conventional Maxwell bridge, on the other hand, is independent of frequency, and the Hay bridge is largely independent of frequency providing the loss tangent is less than 0.1 .

As with capacitance measurement, the loss in an inductance is indicated in terms of effective shunt resistance by the transformer-ratio-arm bridge. Where the loss is small this is unimportant; but inductors frequently have a noticeable loss tangent, especially at low frequencies. At low frequencies the loss in air-cored inductors is principally due to the resistance of the conductors and is therefore effectively series resistance. This must be taken into account when determining the inductance, because, for loss tangents greater than 0.1 , the measured inductance with effective shunt loss may be appreci-
ably different from that with effective series loss. This is explained fully in the appendix to this article.

Three-terminal measurement of inductance at low frequencies is useful mainly for measurement of components in situ. With low-value inductors the impedance of any shunt strays is unlikely to be low enough to produce any noticeable error, the main source of error being the series inductance of the connecting leads. With high-value inductors the self-capacitance of the inductor itself is usually a more important source of error than stray shunt capacitance between the connecting leads or bridge terminals.

Resistance Measurement.-Being essentially an a.c. measuring device, the transformer-ratio-arm bridge measures both the reactance and the resistance of any impedance. This has no particular advantages or disadvantages; but it will be appreciated that " universal" or " LCR" versions of transformer-ratio-arm bridges are switched for capacitance or inductance measurement, and there is no condition equivalent to the "Wheatstone-bridge" setting of the conventional universal bridge.

Bridges having measuring accuracies better than $0.5 \%$ are normally multi-decade bridges (see Fig. 7) and are necessarily calibrated in terms of conductance. Simple resistance measurement is thus less direct than with conventional bridges. Where the measuring accuracy is of the order of $1 \%$, however, a continuously-variable standard can be used having direct calibration.

For measurement of very-high resistance values the transformer-ratio-arm arrangement has the advantage that shunt leakage resistance can often be cancelled by three-terminal measurement.

## APPENDIX

Effective Series and Shunt Loss in Reactive Components.-A series impedance combination $\mathrm{R}_{1}+\mathrm{j} x_{1}$, in one arm of a bridge can be balanced by a shunt impedance combination, $\mathfrak{j} x \mathbf{R} /(\mathbf{R}+\mathfrak{j} x)$, in the opposite arm. Assuming the ratio arms are equal, conditions for balance are:-

$$
\begin{aligned}
& \mathrm{R}_{1}+\mathrm{j} x_{1}=\mathrm{j} x \mathrm{R} /(\mathrm{R}+\mathrm{j} x) \\
& \text { so that, } x_{1}=\frac{x}{1+x^{2} / \mathrm{R}^{2}}=\frac{x}{1+\tan ^{2} \delta} \\
& \text { and, } \mathrm{R}_{1}=\frac{\mathrm{R}}{1+\mathrm{R}^{2} / x^{2}}=\frac{\mathrm{R}}{1+\mathbf{Q}^{2}}
\end{aligned}
$$

If R is large compared with $x$-i.e., the loss tangent is small:-

$$
x_{1}=x, \text { and } \mathrm{R}_{1}=\frac{1}{\frac{1}{\mathrm{R}}+\mathrm{R}}
$$

which is approximately equal to $1 / R$.
Thus, when the loss is small, it is relatively unimportant whether the bridge measures in terms of equivalent shunt or series resistance. For loss tangents greater than 0.1 however, the error in measured reactance becomes comparable with the accuracy of the bridge, and a correction should be made if it is possible to determine whether the unknown has series or shunt loss.

The table gives correction factors to be
applied when an impedance having series loss is measured with a transformer-ratio-arm bridgethe indicated inductance or capacitance should be multiplied by the correction factor.

| $\operatorname{Tan} \delta(x / \mathbf{R})$ | Correction Factor |  |
| :---: | :---: | :---: |
|  | Inductance | Capacitance |
| 0.01 | 1.000 | 1.000 |
| 0.02 | 1.000 | 1.000 |
| 0.04 | 1.000 | 1.002 |
| 0.05 | 1.000 | 1.004 |
| 0.08 | 0.990 | 1.006 |
| 0.10 | 0.990 | 1.010 |
| 0.15 | 0.980 | 1.023 |
| 0.20 | 0.962 | 1.040 |
| 0.25 | 0.943 | 1.063 |
| 0.30 | 0.917 | 1.090 |
| 0.33 | 0.901 | 1.109 |
| 0.35 | 0.893 | 1.123 |
| 0.40 | 0.862 | 1.160 |
| 0.45 | 0.833 | 1.203 |
| 0.50 | 0.800 | 1.250 |
| 0.60 | 0.735 | 1.360 |
| 0.70 0.80 | 0.671 | 1.490 |
| 0.80 | 0.610 | 1.640 |
| 0.90 | 0.553 | 1.810 |
| 1.00 | 0.500 | 2.000 |

Parametric Amplifier using only clectrostatic fields to produce focusing and interaction of an electron beam is proposed by B. J. Udelson in a letter in the August 1960 issue of Proc.I.R.E. (In previous parametric amplifiers using electron beams, as well as electrostatic fields, magnetic fields have had to be used, either to focus the beam or, as in Adler-type valves using cyclotron resonance, to produce interaction.) In the new proposed valve the electron beam would pass between a linear array of pairs of parallel plates divided into input, pump and output sections and fed alternately from two different direct voltage sources. The voltage differences between adjacent pairs of plates produce forces which both focus the beam and also give rise to transverse oscillations of the electrons (at a frequency determined by the voltages and dimensions of the plates). These transverse oscillations can be made to interact with a signal at the same frequency and pump at twice their frequency by superimposing the signal and pump voltages on to the direct parallel plate voltages. Low-noise fast-wave interaction should be possible in such valves.

Permanent Marking of Teflon-insulated wires is difficult because the insulation is one of the least-easily damaged materials used. However, a process for the production of identifying markings which are as durable as the insulation itself has been developed by the Martin Company of Baltimore, Maryland, U.S.A. Teflon foil, bearing identifying letters or figures of darker colour than the insulation, is placed on the wire which then passes into an infra-red oven. The dark legend heats up rapidly and "sinters" on to the insulation, forming an indelible marking, whilst a fan draws in cool air to prevent the air temperature rising to the point where the whole covering is melted.

Interstellar Communication possibilities using optical masers are discussed by Dr. R. N. Schwartz and Prof. C. H. Townes in an article in the April 15, 1961, issue of Nature. Because of the intense beams available from such masers, such communication seems reasonably possible. The best chance of producing a signal distinguishable from the background light of the local star would be to transmit within one of the absorption spectral lines of this star, since the stellar intensity is re-
duced by a factor of 10 or more in such lines, The authors therefore suggest a search for very narrow emission lines (such as would be produced by optical masers) within the absorption lines of the spectra of neighbouring stars. (Possibilities of using radio waves for interstellar communication were discussed by G. Cocconi and P. Morrison on p . 844 of the Sept. 19, 1959, issue of Nature and in this section of our February 1960 issue.)

Computer-Designed Computer has been recently constructed using the Bell Laboratories Automatic Design System (abbreviated to BLADES). In this system the computer is designed as far as possible around a single general-purpose type of logic unit. (Each such unit contains four independent switching circuits which can be interconnected in different ways to produce various types of logic circuit.) The required computer functions are expressed in logical and geometrical terms, which are then used to programme an ordinary gen-eral-purpose digital computer. This then specifies the number of generalpurpose logic units required for each sub-assembly of the computer being designed, the logic unit pins to be interconnected, the sizes of wire to be used in connecting them, the connecting wire paths to be followed for minimum path length, any special logic units required, and finally a complete parts list. Power wiring is also arranged so that no two successive functions are supplied from the same power line.

Magnetic Tape Tester used by M.S.S. uses a high-quality tape recorder to record signals at 1 and $10 \mathrm{kc} / \mathrm{s}$ on the tape. These signals are replayed through the same recorder, and the replay output rectified and displayed on a c.r.t. This enables any fluctuations in the tape output to be observed. Both the rise and fall times of the rectified output display are made independently variable to allow different phenomena to be observed under the best conditions. For example, for observing modulation noise (which can be very "peaky") fast rise and slow fall response times are required, whereas just the opposite-slow rise and fast fall response times-are required for observing short dropouts.

New Magnetic Tape introduced by Gevaert-Type F -has been given a rough backing so as to reduce the risk of spillage when using large
professional open- and single-sided spools. The roughness is produced in this tape by applying a coating to the normal tape backing, rather than by mechanically roughening the backing itself.

Audio-Frequency "Comb" Filter described by L. H. Bedford in the April 1961 issue of Electronic Technology consists simply of a stretched (non-magnetic) conductor, the centre portion of which lies in a transverse magnetic field. At frequencies at which the stretched conductor is mechanically resonant, its electrical impedance rises. This rise can be made use of, for example by connecting the string in a feedback path, to produce a "comb" filter, i.e. a filter having pass or stop bands spaced at uniform frequency intervals. Quite high Q's can be achieved in this way; for example in the phosphor bronze wire $50 \mathrm{c} / \mathrm{s}$ fundamental filter for reducing hum described in the article, the fundamental $Q$ was 220 . At harmonic resonant frequencies even greater Q's (of the order of 1500) were observed. This last factor usefully counteracts the reduction in the attenuation at exact harmonic frequencies which would otherwise be caused by the small (of the order of $0.4 \%$ ) departures of the mechanical harmonic resonant frequencies from a true harmonic series. Resonant attenuations of the order of 20 dB were observed.

New Ultrasonic Transducer being developed at the Bell Telephone Laboratories consists simply of a plate of a piezoelectric semiconductor (such as gallium arsenide) on which a thin metal film is deposited. Since the charge carriers in these two materials have different Fermi quantum energy levels, an electric field is produced across their junction. This field decreases the density of charge carriers around this junction, producing a highresistivity region which is known as a depletion layer. If a voltage is now applied across the transducer, most of this voltage is dropped across the high resistivity depletion layer, so that this layer behaves as a very thin $\left(10^{-5}-10^{-3} \mathrm{~cm}\right.$ thick) piezoelectric crystal. Because of this extreme thinness large piezoelectric stresses are produced, and the efficiency is greatest at very high frequencies. (In the 1 to $10 \mathrm{kMc} / \mathrm{s}$ range, efficiencies as high as 100 times those of other known ultrasonic transducers should be achievable.) Another advantage of the new
type of transducer is that, by varying a direct bias voltage across the transducer, the thickness of the depletion layer and thus also the transducer resonant frequency can be varied: this is not possible with normal piezoelectric transducers. Bandwidths of the order of $5 \%$ or more, i.e., considerably more than at present obtainable, should also be possible. The main use of the new transducer is expected to be to allow longer delay lines to be produced since the higher efficiency will make up for the greater delay-line losses. The longer delay and greater bandwidth should allow a considerably greater amount of information to be stored in such lines. Work at higher frequencies should also be possible with the iew transducer: although still in the early research stage, operation up to $830 \mathrm{Mc} / \mathrm{s}$ has already been achieved.

Character Readers using an area covered by a number of small detectors are not, of course, a new idea. An unusual feature of such a reader developed by the Electrical Engineering Department of the University of Birmingham, however, is that the outputs from the detectors (in this case photo-diodes) instead of being individually processed in logic circuits, are summed in a resistive matrix to give an output level corresponding to the character being read. In this way the interfering effects of noise (character imperfections) can be greatly reduced.
"Potting" of sub-assemblies in polyester or epoxy resins is often carried out to protect the components and wiring from deterioration due to vibration or moisture. Heat is another potential source of danger though, whether applied externally as in the case of an exothermic reaction of the potting resin or internally from electrical dissipation within the components themselves.

Croxton and Garry Ltd., of High Street, Kingston-on-Thames, Surrey, are the sole licensees of a process employing an aluminium wool which, when added to the "mix," increases considerably the heat conductivity of potting resins.
Typically, heat conductivities of the treated resin range from 20 B.Th.U. $/ \mathrm{hr} / \mathrm{ft}^{3} / \mathrm{ft} /{ }^{\circ} \mathrm{F}$ for added aluminium wool $30 \%$ of the weight of the total mix, to 100 B.Th.U./hr/ $\mathrm{ft}^{2} / \mathrm{ft} /{ }^{\circ} \mathrm{F}$ for just over $90 \%$; botll results being obtained with a pure resin of conductivity 0.13 . These values are far better than those obtained either by the addition of steel wool ( $90 \%$, conductivity about 15) or aluminium powder (negligible effect until almost $100 \%$ concentration).

Naturally, electrical conductivity is also increased by the added metal and an insulating skin may have to be used between the components and aluminium loaded resin.

## CONFERENCES AND EXHIBITIONS

Latest information on forthcoming events both in the U.K. and abroad is given below. Further details are obtainable from the addresses in parentheses.
LONDON
June 12-16 Components and Materials used in Electronic Central Hall Engineering (Conference)
(I.E.E., Savoy Place, W.C.2.)

June 19-22 Laboratory Apparatus and Materials Horticultural Hall (U.T.P. Exhibitions, 9 Gough Sq., E.C.4.)

June 21- International Plastics Exhibition and Convention Olympia July 1 (British Plastics, Dorset House, Stamford Street, S.E.1.)
June 26-28 European Symposium on Space Technology 21 Tothill Street (British Interplanetary Society, 12 Bessborough Gardens, S.W.1.)
July 7-29 Soviet Trade Fair (Industrial \& Trade Fairs, Russell Street, W.C.2.)
Aug. 23- National Radio and Television Show Earls Court Sept. 2 (Radio Industry Exhibitions, 59 Russell Sq., W.C.1.)
Sept. 6-8 Microwave Measurement Techniques Conference Savoy Place (I.E.E., Savoy Place, W.C.2.)

Sept. 13-15 Symposium on Photomultiplier Tube Manchester Sq., W. 1 Applications
(E.M.I. Electronics, Valve Division, Hayes, Miaax.)

Oct. 4-12 Computer Exhibition and Symposium Olympia

|  | Nov. 8-10 | Non-Destructive Testing in Electrical <br>  <br>  <br>  <br> (I.E.E., Savoy Place, W.C.2.)$\quad$ Savoy Place |
| :--- | :--- | :--- |

FARNBOROUGH

| Sept. 4-10 | Farnborough Air Show <br> (S.B.A.C., 29 King Street, London, S.W.1.) |
| :--- | :--- |
| NORWICH |  |
| Aug. 30- <br> Sept. 6 | British Association for the Advancement <br> of Science annual meeting <br> (B.A.A.S., 19 Adam Street, London, W.C.2.) |
| OXFORD | Communications and Space Research Convention |
| July 5-9 | (Brit. I.R.E., 9 Bedford Square, London, W.C.1.) <br> Engineering Inspection and Non-Destructive Testing |
| Sept. 2-8 | (Institution of Engineering Inspection, 616 Grand <br> Bldgs., Trafalgar Sq., London, W.C.2.) |
|  |  |

OVERSEAS
June 12-25
International Exhibition of Electronics, Nuclear
Energy, Radio, Television and Cinematography
(Fairs \& Exhibitions, 2 Dunraven St., London, W.l.)
June 26- International Measurement Conference Budapest
July
(Prof. J. F. Coales, The University, Cambridge.)
Joint Automatic Control Conference
(Dr. R. Kramer, M.I.T., Cambridge 39, Mass.)
July 9-14
July 16-21
Aug. 22-25
Aug. 25-
Sept. 3
Sept. 1-8
Sept. 1-10
Sept. 11-15
Bio-Medical Electronics Conference
(Dr. A. Rémond, 131 Boulevard Malesherbes, Paris 17.)
Medical Electronics Conference
(Dr. A. Nightingale, St. Thomas's Hospital, London, S.E.1
(Dr. A. Nightingale, St. Thomas's Hospital, London, S.E.1.)
Western Electronics Show and Convention San Francisco
(Wescon, 1435 LaCienega Blvd., Los Angeles.)
German Radio Exhibition
(Berliner Ausstellungen, Charlottenburg 9, Berlin.)
Firato-International Radio Show
Berlin

Amsterdam
(Firato Secretariat, Emmalaan 20, Amsterdam.)
Danish Radio Exhibition
Copenhagen
International Cybernetics Congress
Namur
(International Association of Cybernetics,
13 Rue Basse-Marcelle, Namur, Belgium.)
Sept. 11-19
International Congress of Navigation
Baltimore
(Permanent International Association of Navigation Congresses, 60 rue Juste Lipse, Brussels, Belgium.)
Sept. 14-25 French Electronics, Radio \& Television Show
Paris
(F.N.I.E., 23 rue de Lubeck, Paris XVI.)

Sept. 20-21 Industrial Electronics Symposium
Boston
(I.R.E., 1 E. 79 St., New York, 21.)

Sept. 25-30 $\quad$ Irish Television and Radio Show
Oct. 2-4
I.R.E. Canadian Convention

Dublin
Toronto
(I.R.E. Convention, 1819 Yonge Street, Toranto 7.)

Oct. 9-11 National Electronics Conference
Chicago
(N.E.C., 228 N. LaSalle St., Chicago.)

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By "DIALLIST"

## A Remarkable TV Feat

THE B.B.C., having made all arrangements for the May Day television transmissions from Moscow, was in the fortunate position of being able to broadcast Yuri Gagarin's reception in Moscow after his space flight. It was a wonderful achievement as was the May Day transmission. Signals travelled over a rather complicated route: Moscow, Leningrad, Tallinn, Helsinki, Stockholm, Copenhagen, Brussels and so to London. Considering the length of their journey and the many stages of reception and retransmission through which they had to pass, the signals gave remarkably good and steady pictures in this country. Now that a link with Moscow has proved successful the accomplishment of round-the-world television could become a reality while we are waiting to see whether satellite communications are practicable. Relays would cross Siberia to somewhere near the Bering Strait and go thence to Canada and the U.S.A. thus linking the two hemispheres. France has already a tropospheric scatter link with Algeria and in time to come Europe and many parts of Africa should be able to exchange programmes. What a wonderful world it'll become when the say-
ing that one half of it doesn't know how the other half lives ceases to be true.

## Inter-school Television

SOMEHOW, I'm personally not at all enthusiastic about the use of TV for teaching purposes. For lectures, yes: classes get a chance of listening to someone eminent in his or her own line which might not otherwise come their way. But there's a world of difference between lecturing and teaching. The former is more or less impersonal: the lecturer tries to put his subject over in a clear and interesting way; but he's little chance of discovering whether or not it's sinking in as it should. Teaching on the other hand is (or should be) a closely personal business. The teacher knows the members of his class and has every opportunity of finding out that what he says is learnt and not just listened to.

## 'Phone Cable Round the World

ON 8th April Monarch, the world's biggest cable-laying ship (owned by Cable \& Wireless), sailed from Oban

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and began to pay out cable for CANTAT, the transatlantic link of what by 1964 will provide the British Commonwealth round - the - world telephone cable system. The other end of this link is Hampden in Newfoundland, 2,100 miles away. As Monarch can'l carry more than 700 miles of cable she'll twice have to buoy the end and return here for another consignment. From Newfoundland the link is to cross Canada partly by cable and partly by wireless to Vancouver. There it will connect with the cable to be laid to Hawaii, Fiji, New Zealand and Australia. The cable now being laid is of an entirely new type, developed at the G.P.O.'s Dollis Hill research station, and made by Submarine Cable, Ltd. It has a steel core and round this is the first conductor, a copper tube. A layer of polythene insulation and then comes an aluminium tube as the second conductor. Then after further layers of insulation and screening a strong polythene outer casing. With the exception of the sections laid in shallow water the cable is not armoured. It is so much more efficient than earlier cables that repeaters will be 26 miles apart instead of 16 . The new cable will provide 60 two-way circuits. An interesting little booklet outlining the "Story of the Submarine Cable" from 1849 to this latest project has been produced by Submarine Cables Ltd. which, as you may know, is owned jointly by A.E.I. and B.I. Callender's Cables.

## Electronic Locks

IN the United States (and for all I know in this country, too) work on the production of electronic locks for safes and strong rooms is going ahead. The bolt is thrust home by an electro-magnet or by a motor and with power of that kind doing the job it can be made to engage much more securely than is possible with a key-operated bolt. Outside the door to which it is fitted is a panel carrying ten numbered buttons, which must be pressed in a predetermined order to actuate the lock. Should would-be burglars try to work the lock and get the combination wrong an alarm signal is radia-
ted. The idea looks pretty good to me and I expect we'll hear much more in days to come of electronic locks.

## "Lifeline"

THE use of inflatable life rafts by ships of all kinds is rapidly increasing. For use in these Marconi's have marketed a special small lightweight transmitter receiver. "Lifeline," as it's called, weighs less than 20 lbs and measures only $17 \times 7 \times 5$ inches. There's a telescopic aerial which is very easily erected. In use, the equipment is strapped to the operator's knec. Only two controls are needed: one selects speech or an autcmatic alarm signal; the other funes the transmitter for the best possible performance. The equipment is pretuned to $2,182 \mathrm{kc} / \mathrm{s}$, the international distress and calling frequency. Much use is made of transistors to keep down both size and power demands. Power is supplied by a generator worked by a handle and it is stated that little effort is needed to turn this.

## Radio Towers

IN a recent number of the Journal of the Institution of Electrical Engineers, J. F. Goodger described the television tower at Stuttgart as a smokestack surmounted by a crow's nest. And that's just what it looks like. In this stack are highspeed lifts which carry visitors to restaurants and viewing galleries 500 ft above the ground. We're told that in the five years since the tower was built four million people have visited it and have more than repaid the cost of putting it up. This makes one wonder whether there'll be similar happenings in London, where a $500-\mathrm{ft}$ radio link tower is to be built at the Museum telephone exchange, and is due to be completed by the end of 1963. It will replace the existing radio mast with which an effective height of only l80ft is obtained. The tall buildings that have gone up in London during the past few years and the many more that are planned make a much greater effective height necessary. It will be used by the Post Office as a radio link for the telephone and television services. Like the Stuttgart tower, it is to have a public viewing platform-at a height of 463 feet. If a moderate admission fee is charged, it, too, should pay for itself within a few years.


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## By "FREE GRID"

## Jiggers and Gee-gees

IN our Jubilee issue the learned contributor of the historical survey drew attention to our ignorance of the origin of the word "jigger," which was used in the early days of wireless to describe the r.f. transformer which linked aerial and closed circuit. I hope to lighten his darkness forthwith, and if I'm wrong I feel sure some of you will put me right.
The word "iigger" had, and still has, certain particularized meanings aboard ship and also ashore but, in general, it means an auxiliary device; it is, in fact, something which carries out a function described by the verbal expression "to jig up," which means to increase or step-up something. Surely this meaning of stepping up or increasing describes precisely the function of the early wireless jigger which did indeed jig up the volts as our contributor indicaton (April issue, page 158).
The expression "to jig up" was, and probably still is, in very common use among certain horse dealers, particularly Irish ones, and when we


Common among horse dealers
recall that Marconi's mother was Irish, and he paid many visits to his Irish grandparents in his boyhood, we begin to see daylight.
For the benefit of non-horsey people, I should explain that the term "jigged up" describes the result of dosing a horse to make it prance in a lively manner, reminiscent of an Irish jig, in front of its intending buyer. Such treatment involves the use of ginger and is sometimes referred to as gingering up.
Without doubt, Marconi, in his boyhood visits to his mother's native land, would have become acquainted
with the expression "jigging up," as applied to horses, and he would also have recognized in the Irish jig, a dance which, in his native Italy, was called giga, pronounced "jeegah."

## Ham and Eggs

MY enquiry in the March issue as to the precise meaning and derivation of the " $x$ " in the well-known abbreviation xtal has provoked some interesting replies. Many correspondents seem to think the " x " is the Greek letter "chi" which is used in Xmas. This seems all wrong because there is no "ch" in crystal as there is in Christmas.
At the same time I dropped a brick when I suggested that the "x" in the common abbreviation xmitter came from the Latin "trans." A German correspondent, Paul Hameyer, to whose excellent English I pay tribute, has made it clear to me that it is a natural abbreviation of the Latin word "ex."
He also convinces me that the " $x$ " in xtal was the work of the American Hams who delighted in such abbreviations as rx for receiver to an even greater extent than their British opposite numbers, who might well have been called Eggs, so closely have the amateur transmitters in the two countries always been linked together.

## Morphean Music

IT is far too late for me to make any comments on the Audio Festival and Fair, but I feel that I cannot refrain from calling attention to the lack of a very necessary audio service in our good-class hotels such as the one in which the Fair was held.
When in the various bedrooms in which demonstrations were given, I could not help noticing that over each basin a socket for an electric shaver was provided, this being of the very latest type with a built-in isolating transformer. But I looked in vain for a socket into which a pillowphone could be plugged so that a sleepless guest might lull himself into dreamland with morphean music.
All-night music is not available from the B.B.C., but surely in these days of tape recorders it is not too much to expect an hotel to provide an all-night service from recorded tapes. No expensive disc-jockey would be needed as in these days of long-playing multi-track recorders, an automatic all-night service could be
provided. By using a three-wire circuit and a simple switch, it would be possible to give the sleepless guest the choice of two programmes; after all, the cost of a couple of tape recorders should be a small matter to an hotel.
Reverting to the electric-shaver socket, there is one curious thing which I observed in this particular hotel which I have also noticed in all others where this facility is provided, and that is that the majority of the guests expected would seem to be monks. The reason is that one bar light is provided, and this is invariably above the mirror where it shines on the top of the head, and could, therefore, only be of real use to a monk desiring to trim his tonsure.
Experience has taught me that a bar light on the top of the mirror is almost useless for shaving, especially under the chin. What I have at home is a bar light on each side of the mirror. Perhaps the various hotel managements would attend to this small point when putting in the wiring for the morphean music service I have suggested. An even better idea to my mind would be that the shaver makers should provide their products with a small floodlight such as is fitted to certain vacuum cleaners. This is, however, really more a matter for our sister journal, Electrical Review, than for Wireless World.

## Photaural Fair

JUST about the time some of you are reading these lines the biennial Photo Fair will be on at Olympia (May 29th to June 3rd). I am not getting any bribe from our sister journal Amateur Photographer-not even a free ticket-for this free advertisement. I only mention it because I always think it a pity it does not join forces with the Audio Festival and Fair because both exhibitions are really for more or less the same type of people.
I find that most people like myself, who take a serious interest in hi-fi audio, are also interested in hi-fi fotos, and most, again like myself, have about the same technical knowledge of each art or science.
At the Photo Fair there are a certain number of inexpensive "snapshot" cameras, but on the whole, the show is not intended for the soot and whitewash brigade any more than the Audio Fair is meant to attract those whose taste in music is the jungle jangling of juke boxes.
If these two fairs put their resources together they could afford to rent the main Olympia hall as well as the lesser ones and so be able to erect properly designed demonstration theatres for both the audio and video arts, and for the best of stereo in both the arts.

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$24 V$
$24 V$
$24 V$
$24 V$
$24 V$
$24 V$

| 15 A | ... | 614 |
| :---: | :---: | :---: |
| 50 A | ...... | ¢47 |
| 100 A | ....... | ¢66 |
| 10 A | ......... | ¢15 |
| 20 A | ......... | ¢22 |
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| 105 A |  | ¢62 |
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| 1,000 A | ......... | ¢185 |
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36 V
36 v
36 v
36 v
110 v
110 v
110 V
110 V
110 V
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250 v
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| :---: | :---: |
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\begin{array}{c|c}
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\hline 50 \mathrm{c} / \mathrm{s} & \text { From } 50 \mathrm{~W} \text { to } 250 \mathrm{~W} \\
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\end{array}
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| PP 4090 X | Multi-hole Directional Coupler | 1.05 | Directivity: $>40 \mathrm{~dB}$, coupling factor 10 dB or $20 \mathrm{~dB}( \pm 0.2 \mathrm{~dB}$ ) |
| PP 4095 X | Cross-guide Directional Coupler | - | Directivity: 20 dB , coupling factor 26 dB ( $\pm 0.5 \mathrm{~dB}$ ) |
| PP 4110 X | Fixed Attenuator | 1.10 | Attenuation 6, 10 or $20 \mathrm{~dB}( \pm 0.2 \mathrm{~dB}$ ), max. peak power I kW |
| PP 4130 X | Variable Flap Attenuator | 1.15 | Max. attenuation $>20 \mathrm{~dB}$ |
| PP $4150 \times$ | Variable Rotary Attenuator | 1.15 | Max. mean power I W, max. attenuation 50 dB , accuracy $\pm 2 \%$ |
| PP 4170 X | Low-power Matched Load | 1.05 | Max. mean power 2 W |
| PP 4200 X | Klystron Mount | - | When using klystron 2 K 25 the output power is $>20 \mathrm{~mW}$ |
| PP 4220 X | Adjustable X-tal Mount | 1.10 | Sensitlvity: I mV D.C. for $0.1 \mu \mathrm{~W}, 50 \Omega \mathrm{~N}$-connector |
| PP $4225 \times$ | Broadband X-tal Mount | 1.50 | Sensitivity: I mV D.C. for $10 \mu \mathrm{~W}, 50 \Omega$ BNC-connector |
| PP $4245 \times$ | Tunable Thermistor Mount | 1.10 | Freq. range $8.2-11 \mathrm{kMc} / \mathrm{s}, 50 \Omega \mathrm{BNC}$-connector |
| PP 4260 X | Calibrated Short Circuit | $>100$ | Accuracy of the displacement 0.02 mm |
| PP $4280 \times$ | Sliding Screw Tuner | from 20 to $>1.02$ | Insertion loss for a VSWR of 20 is $\mathbf{2} \mathbf{2 d B}$ |
| PP 4290 X | Direct Reading Wavemeter | - | Freq. range $8.5-9.8 \mathrm{kMc} / \mathrm{s}$, absolute accuracy $\pm 2 \mathrm{Mc} / \mathrm{s}$, loaded $Q 10,000$ |
| PP 4300 X | Broadband Wavemeter | 1.10 | Relative accuracy $3.10^{-4}$, loaded $\mathrm{Q}>3.000$ |
| PP $4360 \times$ | Measuring Cavity | - | Freq. range $8.65-8.95 \mathrm{kMc} / \mathrm{s}$, loaded $\mathrm{Q}>3,000$, magnetic field for electron resonance 3,300 gauss |
| PP 4380 X | Standing Wave Detector | 1.05 | Accuracy of the probe displacement 0.01 mm , $50 \Omega$ BNC-connector |
| PP 4385 X | High Precision Standing Wave Detector | - | Measurable VSWR between $1.005-2.000$, accuracy probe displacement $2 \mu$, probe penetration 0.3 mm |
| PP $4421 \times$ | Ferrite Isolator | 1.15 | Freq. range $8.5-9.6 \mathrm{kMc} / \mathrm{s}$, forward att. $<.0 .8 \mathrm{~dB}$, reverse att. $>13 \mathrm{~dB}$, max. peak power 50 kW |
| PP $4422 \times$ | Ferrite Isolator | 1.20 | Freq. range $8.5-9.6 \mathrm{kMc} / \mathrm{s}$, forward att. $<0.5 \mathrm{~dB}$, reverse att. $>20 \mathrm{~dB}$, max. mean power 1 W |
| PP $4500 \times$ | $3 \mathrm{~cm} \cdot$ Noise Generator | > 1.20 | Nolse factor 18.7 dB ( K 50 A ), attenuation 0.13 dB |

Additional Instruments: D.C. Microvoltmeter, type GM 6020. Klystron Supply, type GM 4561. Bolometer Bridge, type GM 4460

| (n) TYPE | DESCRIPTION | MAX. VSWR. |  |
| :---: | :---: | :---: | :---: |
| PP 4020 Q | Straight Waveguide Section | - | Length 5, 10 or 20 cm |
| PP 4025 Q | E-plane Bend | 1.07 | Radius of curvature 35 mm |
| PP 4030 Q | H-plane Bend | 1.07 | Radius of curvature 37 mm |
| PP 4035 Q | Twist | 1.07 | Length 50 mm |
| PP 4050 Q | Hybrid Tee | I5 | Decoupling $>35 \mathrm{~dB}$ |
| PP 4080 Q | Horn | 1.15 | Directivity: E-plane $15^{\circ}$, H-plane $16^{\circ}$ |
| PP 4130 Q | Variable Flap Attenuator | 1.15 | Max. attenuation $>20 \mathrm{~dB}$ max. mean power 200 mW |
| PP 4150 Q | Variable Rotary Attenuator | 1.15 | Max. attenuation 50 dB , accuracy $\pm 3 \%$ |
| PP 4170 Q | Low-power Matched Load | 1.05 | Max. mean power I W |
| PP 4200 Q | Klystron Mount | $\overline{7}$ | When using klystron 55,335 the output power is 100 mW |
| PP 4222 Q | Adjustable X-tal Mount | 1.25 | $50 \Omega$ BNC-connector |
| PP 4260 Q | Calibrated Short Circuit | 50 | Accuracy of the displacement 0.02 mm |
| PP 4270 Q | Sliding Screw Tuner | from 10 to 1.03 | Insertion loss for a VSWR of 10 is $\geqslant 2 \mathrm{~dB}$ |
| PP 4300 Q | Broadband Wavemeter | 1.20 | Relative accuracy $5.10^{-4}$ loaded $Q>3000$ |
| PP 4382 Q | Standing Wave Detector | 1.03 | Accuracy of the displacement 0.01 mm max. probe penetration $1 \mathrm{~mm}, 50 \Omega \mathrm{BNC}$-connector |
| PP 4420 Q | Ferrite Isolator | 1.15 | Freq. range $33-36 \mathrm{kMc} / \mathrm{s}$, forward att. $<1 \mathrm{~dB}$ reverse att. $13-26 \mathrm{~dB}$, max. mean power 200 mW |

[^10]
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|  <br>  | Typical Peak Out ut Power (kW) |
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|  <br>  Aむ | Magnetron |
|  | Typical Peak Output Power (kW) |
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ARTICLES
IN THE JUNE ISSUE INCLUDE:

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produced by the action of light are emitred from the sensitive profaced by the action or inhr are cmitred from the sensitive surface into a accuum, or near vacuum, eo be coilected by ed positive electrock. Some aspects of human ins theory, and of in photomectyy, cementary physics of quantum hnery, the photoclectric process are includec.
printed-circuit plating
The noble metal plating of printed circuits is discussed in this article. Details are given of the various metals used and their properties are consid dered. Tables are included which indicate recommended plating thicknesses for most modern materials along with illustrations and details of plating techniques.

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[^12]
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[^13]
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 477

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Belozs is reproduced the response curve of the AS 7012 which is typical of the whole of the Solent Series.


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ABRIDGED DATA
$\mathrm{Vh} \quad=6.3 \mathrm{~V}$
$\nabla \mathrm{Fh}-\mathrm{k}(\mathrm{pk}) \quad \mathrm{max} .= \pm 250 \mathrm{~V}$
$\nabla a_{1}+a_{3} \quad$ min. $=350 \mathrm{~V}$
$\mathrm{Sx} \quad=\frac{95}{\mathrm{~V} \mathrm{a}_{3}} \mathrm{~mm} / \mathrm{V}$
Sy $\quad=\frac{110}{V a_{3}} \mathrm{~mm} / \mathrm{V}$
(The green medium persistence phosphor used by E.T.L. has now been designated
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Gap $.0002^{\text {R/P }}$
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. 55 Inductance at I Kc. 7.5 MV. Output at I Kc. Fully Modulated Tape.

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Dual gap.
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| encapsulated in Epoxy Resin <br> 4. Metal Oxide Precision | $\begin{gathered} 1 / 4,1 / 2,1 \\ 1 / 4-2 \end{gathered}$ | $\begin{gathered} 10 \Omega-4.7 \mathrm{M} \\ 100-4 \cdot 2 \mathrm{M} \end{gathered}$ | $\begin{aligned} & 1 \% 2 \% 5 \% \\ & 1 \% 2 \% 5 \% \end{aligned}$ |
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VERTICAL-DEFLECTION SYSTEM
MAIN FRAME AMPLIFIER
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DC coupled to $36 \mathrm{Mc} / \mathrm{s}$.
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0.009 microsecond ( 9 nsec ).

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Calibrated Sweep Rates From 0.05 microsecond $/ \mathrm{cm}$ to 2 seconds/cm in 24 steps.
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## Pye at Dounreay

The Pye Instrument Group has supplied all the equipment to the U.K. Atomic Energy Authority for the irradiated fuel element laboratory at Dounreay. In addition to supplying equipment, Pye Ltd. acted as consultants and designers on all matters in that laboratory relating to instrumentation and remote handling. The illustration above shows manipulators working in conjunction with a television camera to handle and measure a sample from the fast reactor.

## " BELLING-LEE " NOTES

No. 29 of a Series
Some mechanical aspects of design: Part 2.
Last month we mentioned that in a multi-pole connector positional tolerances of the poles might result in excessive mating forces and wear. This is particularly true of connectors which have rigidly mounted pins and sockets, and in better class components it is usual to find that they are allowed a certain amount of individual float so that they can take up correct relative alignment. This principle of positional adjustment is used in "Belling-Lee" unitors, and even though the mouldings on which the contacts are mounted are made with extremely high precision, the plug pins are floating and capable of a controlled amount of lateral movement and variation in angle of approach. Notice that we say "controlled", because excessive float could be more harmful than a slight amount of mis-alignment, and might result in fracture in the event of jamming.

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# Aspects of design 

This is No. 35 in the series of articles dealing with advanced problems in circuit design published by The Ediswan Mazda Applications Laboratory. No. 36 will appear next month. We shall be pleased to answer queries arising from this or other articles.
Reprints of the first twenty-four articles, in booklet form, are available on request.

35 OPERATING
CONDITIONS OF
FIELD SCANNING
OUTPUT STAGES*
(PART I)

## *In accordance with BS204: 1960 the internationally used term "field" replaces the previously used term "frame."

## VOLTAGE MEASUREMENTS

When designing output stages for field scanning in television receiver's it is necessary to ensure that the working conditions are such as to provide reliable operation and good valve life. This means that the stage must be capable of providing sufficient power to scan the full height of the tube face in spite of production variations in components and normal variations in supply voltage. In addition, normal production variations in valve characteristics must be taken into account and a safety factor allowed to accommodate sufficient deterioration of characteristics in use to ensure that early replacement is not necessary.

The purpose of the present article is to outline methods of making the necessary voltage measurements to determine the operating conditions of the stage. Current measurements will be the subject of "Aspects of Design No. 36."

## MINIMUM ANODE VOLTAGE

For any field output valve in the Ediswan Mazda range examples are given of available maximum anode current for certain values of minimum anode voltage at the end of the scanning stroke. These figures represent the worst valve likely to pass the factory test, as well as a further deterioration to represent a useful length of life.

It was pointed out in "Aspects of Design No. 3" that the design of the output stage must be carried out in such a way that the anode voltage at the end of the scanning stroke does not swing below this stated minimum even under the most adverse conditions. Accordingly


FIG.I b the measurement of operating conditions should be carried out using the highest value of deflector coil resistance likely to be encountered allowing for normal production variations, with this value further increased to correspond to the highest operating temperature.
The HT line voltage should be the lowest value obtained over the range of mains taps on the receiver (with on the receiver (with but this value should be further reduced to allow for variations in supply voltage.

Fig. 1 shows the circuit of a typical field output stage and Fig. 2 shows a typical anode voltage waveform. The presence of the high peak voltage during flyback makes accurate measurement of minimum anode voltage difficult. The flyback peak is usually of the order of 1,000 volts. It has been found convenient to carry out minimum anode voltage measurement by means of the circuit shown in Fig. 3. Point $P$ is connected to the anode of the output valve, Q is connected to an oscilloscope of at least 1 megohm input impedance and point $S$ is connected to a d.c. supply of at least 100 volts. With the potentiometer slider at the low potential end, the diode is non-conducting. The output valve anode waveform is displayed on the oscilloscope which

## Associated Electrical Industries Ltd <br> Radio and Electronic Components Division Technical Service Department

 Tel: GERrardi9797? Grams: Sieswan Westcent:Londonmay be of the a.c. coupled type but should have a good response down to at least 50 c.p.s. Resistor $\mathbf{R}$ must be chosen so that, in conjunction with the input impedance of the oscilloscope there is negligible attenuation of the waveform. For this purpose R should have a value of about $2 \%$ of the input impedance of the oscilloscope.
If the potentiometer slider is moved towards a more positive voltage there comes a time when this voltage equals the minimum anode voltage of the valve. The diode then begins to conduct and clipping of the negative extremity of the waveform is seen on the oscilloscope.
When clipping just commences the minimum anode voltage can be read on the voltmeter V . The resistor $R$ is included to increase the impedance of the source from which the voltage waveform is supplied to the clipping diode, thus making the onset of clipping by the diode more noticeable.

fig. 2

Some care must be exercised in the choice of a suitable diode for this application. During the flyback interval, voltage peaks of the order of one thousand volts appear on the output valve anode. The diode must therefore be capable of withstanding peak inverse voltages of this magnitude. In addition the diode must have low impedance in the forward direction for effective clipping. The heater supply may be provided by an isolated winding with insulation suitable for about 2,000 volts operation, in which case heater and cathode may be connected together. The alternative solution which is probably simpler, is to use a diode having a high heater-cathode voltage rating so that a $0 \quad 5$ heater supply at low


It should be emphasised that the figure to be determined in examining the operation of the stage is the minimum anode-tocathode voltage of the valve. Therefore, if this circuit is used to measure minimum anode voltage with respect to chassis, the end-of-scan cathode voltage must be subtracted from the result.

## PEAK ANODE VOLTAGE

It is also necessary to ensure that the Maximum Peak Positive Anode Voltage rating of the valve is not being exceeded. The Peak Positive Anode Voltage can best be measured by using a capacity compensated potentiometer together with a measuring oscilloscope. The potentiometer must be accurately calibrated after compensation, with the oscilloscope connected during calibration and compensation.

## NEW HIGH POWER TRIODE BEAM TETRODE

EDISWAN MAZDA 30PL14
The 30PL14 Triode Beam Tetrode is intended principally for use in frame deflection circuits of television receivers using $110^{\circ}$ and $114^{\circ}$ cathode ray tubes. The terrode section has been designed to operate with a low ratio of screen to anode current whilst the available peak anode current and anode dissipation are higher than in the 30PL13. The higher available peak anode current will, in many cases, permit economies in output transformer design. The triode section is a general purpose triode with identical characteristics to the $6 / 30 \mathrm{~L} 2$, and that of the 30PL13, for use as a deflection drive voltage oscillator, frame sync. pulse separator, etc.

$$
\begin{array}{lll}
\text { Heater Current (amps) } & \mathbf{I}_{\mathrm{h}} & 0.3 \\
\text { Heater Voltage (volts) } & \mathrm{V}_{\mathrm{b}} & 16
\end{array}
$$

## TENTATIVE RATINGS AND DATA

## Maximum Design Centre Ratings



## TRIODE CHARACTERISTICS

| Anode Voltage (volts) | $\mathrm{V}_{\text {a }}$ | 200 |
| :---: | :---: | :---: |
| Anode Current (mA) | $\mathrm{I}_{3}$ | 10 |
| Grid Voltage (volts) | $\mathrm{V}_{\mathrm{gi}}$ | - -7.7 |
| Mutual Conductance (mA/V) | gm |  |



## TETRODE OPERATION IN FRAME TIME BASE

Allowance must be made in circuit design, not only for component variation, but for valve spread and deterioration during life. Values of total tetrode peak anode current, for an average valve when new and at the assumed end of life point for any valve, are as follows:-

|  | $\mathrm{V}_{\mathrm{s}}$ | $\mathrm{V}_{\mathrm{g} 2}$ | $\mathrm{~V}_{\mathrm{g} 1}$ | $\mathrm{I}_{\mathrm{s}}$ |
| :--- | :--- | :--- | :--- | ---: |
|  | $(\mathrm{V})$ | $(\mathrm{V})$ | $(\mathrm{V})$ | $(\mathrm{mA})$ |
| Average New Valve | 55 | 170 | -1 | 210 |
| Assumed End of Life Condition | 50 | 170 | -1 | 135 |
| Average New Valve | 55 | 185 | -1 | 235 |
| Assumed End of Life Condition | 50 | 185 | -1 | 151 |

Tentative Characteristic Curves of Ediswan Mazda Valve Type 30 PL14



Mounting Position: Unrestricted Base: B9A (Noval)

## Connections



VIEW OF FREE ERD


Maximum Dimensions (mm)

| Overall Length | 78.5 |
| :--- | :--- |
| Seated Height | 71.5 |
| Diameter | 22.2 |

Seated Height
71.5

Diamerer


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Any stabilized output voltage in the range 200-250 v. can be selected by plug-board tappings. The selected output voltage is automatically maintained constane within $\pm \frac{1}{2} \%$, at all loads 0 to $30 / 37 \frac{1}{2}$ amps., when the supply ovitage is varying over the range $+8 \%$ to $-12 \%$.
Frequency compensated $45-55$ and $54-66 \mathrm{c} / \mathrm{s}$.
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## NEW! 10,000 O.P.V. MULTI-TESTER ON BOTH AC \& DC <br> \section*{FULL SCALE RANGES:} <br> MODEL EP-IOK OUTPERFORMS INSTRUMENTS MANY TIMES ITS SIZE

D.C. VOLTS; 0-6-30-120-600-1,200 v A.C. VOLTS: $0-6-30-120-600-1,200 \mathrm{v}$. D.C. CURRENT: $0-120$ UA, $0-12-303$ mA. AND PRICE!
UNBELIEVABLE BARGAIN!
A revolutionary new Multi-Tester. A complete wired and tested instrument RESISTANCE: $0-20 \mathrm{~K}, 0-2 \mathrm{Meg}$. ( 150 ohm, 15K at centre scale). CAPACITANCE: 0.005 to 0.15 uF (at A.C. 6 v.).

DECIBELS: -20 to +63 db ( 600 ohms, $1 \mathrm{~mW} 0 \mathrm{dbm}=0.775 \mathrm{v}$.
ACCURACY: D.C. voltage and current $\pm 2 \%$ f.s. A.C. voltage $\pm 4 \%$ f.s. Re(nor a kit) incorporating extra large $3 \frac{1}{2}$ in meter face and unique slide range switch. Can be conveniently carried in the pocket and teatures unusually sensitive 10,000 ohms per volt A.C.-D.C. meter, I per cent precision resistors, and largest meter ever placed on an instrument this size. Single, easy to use range selector switch, can be appreciated by the novice and engineer alike. Complete with colour coded test leads and battery.
Size: $4 \frac{1}{2} \times 3 \frac{3}{2} \times$ lin. Model EP-IOK. ONLY $\$ 5.19 .6$. P. \& P. $3 / 6$


HI-FI HEADPHONES Uses high quality permanent magnetic speakers with regular voice coil. The padded chamois ear-muffs give correct spacing for optimum acoustic load, giving finest music and voice reproduction. WIRELESS SET No. 19
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ONLY 65/-. Carr. 10/-

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## MEGAPHONE MODEL PM. 242

HAND-HELD PORTABLE SOUND BROADCASTER WEIGHING ONLY 4lbs.
A lightweight new megaphone notable for its extreme economy in battery power despite its high sound volume output. Incorporates a patented non-linear current-limiting device to give maximum postalk. The reflexed air column in the horn, plus its special shape serve ta concentrate and direct the amplified sound and throw it for a conoidercin distance. Sturdy construction throughout with lightweight iderable distance. spun aluminium rubber abbed instrument to give the added feature of remote operatio. Eire Depart recommended ments, Railway Yards, Sports Events, Coachica,
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30 watts. Designed on
entirely new principle
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## stable heat character

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TERMS: C.W.O. or C.O.D. No C.O.D. under $E 1$. Postage $1 / 9$ extra on all orders under $£ 2$, 2/9 extra under $£ 5$ unless carriage stated. Trade supplied. Post order to: Mail Order Dept.

HI-FI 10 WATT AMPLIFIERS
BRAND NEW CARTONED
MANUFACTURERS DISCONTINUED
$\mathbf{8 6 . 1 9 . 9}$ MANUFACTURERS DISCONTINUED MODEL. A REMARKABLE OPPORTUNITY. Carr. 76. Dual separately controlled inputs, for mike and gram.
Separate base and treble controls. HIgh sensititity output for 3 ohm or 15 ohm loudspenker. Guaranteed, tested and in perfect working order. Please state speaker matching required when ordering.

## SUPERHET RADIO FEEDER UNIT

 Design of a high quality Radio Tuner Unit (specialiy suitablefor uue with any of our Amplifers). A Triode Heptode Fichnnger is used. Pentode I.F. and double Diode second Detector delayed A.V.C. is arranged so that A.V.C. dis-
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9.6 .7 in . high. send 8.A.E. for illustrated leaflet. Total 9.6.7in. high. \$end $8 . A . E$. for illustrated leaflet. Total
building cost is $£ 4 / 15 /=$. Polnt to-Polat wiring diagrams
R.S.C. HI-FI TAPE RECORDER KIT

Bulld a high quality recorder in the $£ 70$ class for only 25를를 OR DEPOSIT $25 / 7 / 6$ and 12
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Cash price if settled in 3 months Can be assembled in $\frac{1}{t}$ hour. INCORPORATING THE LATEST COLLARO STUDIO TAPE TRANSCRIPTOR 7x 4in. LOUDSPEAKER. Reel of Best Quallty TAPE, Spare Tape Spool, a Portable Cabinet, size approz. 16 天 13 g 9in, finished to du rable snd atiractive duo-
tone Policrome and comenection diagram lor wiring ampliter to transcriptor. FEATURES INOLUDE:
$\star 3$ SPEEDS $\star$ FREQUENGY RESPONSE $50-11,000$ o.p.s. $\star$ SWITCEED NEGATIVE FEEDBACK EQUALIZING FOR EACH SPEED. $\star$ OUTPUT 4 WATTS $\star$ MAIC EYE RECORDNG LEVEL INDICATOR $\star 3$ MOTORS
Fsit rewind $\star$ TAPE MEASURING AND GALBRATIG DEVCE $\star$ TAKES
 effective adtomatic erasure.
Full descriptive leaftet supplied on receipt of S.A.E.


## R.S.C. BATTERY TO MAINS CONVERSION UNITS

Type BM1. An all-dry battery ellminator. Size 5$\} \times 41 \times 2$ 2in. approx. Com-
pietely replaces batteries supply 1.4 v . and $90 \mathrm{\nabla}$. where A.C. mains $200-250 \mathrm{~F}$.
$50 \mathrm{c} / \mathrm{s}$ is avallable. Suitable for all battery portable receivers requiring 1.4 a and 90 r . This includes latest low consumption types. Complete kit with diagram $39 / 9$ or ready for use $48 / 9$.
 BOTH 0.4 a. to 1 amp., fully smoothed. THEREBY COMPLETELY REPLACING BOTA E.T. BATTERIES AND L.T. 2 v. ACCUMOLATORS when connected
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RECEIVERS normally using 2 v. accumulator.
Complete kit with diagrams and instructions. $49 / 9$ or ready for use $59 / 6$
POWER PACK KITS. Only 19/11. Fully smoothed H.T. output of 250 v 60 mA and L.T. supply of 0.3 v. 1.5 amp . Consistlag of Double Wound Mains
 Ing Choke, Double Electrolytic Condenser, Aluminium Chass is and Circuit.

## R.S.C. A12 STEREO AMPLIFIER KIT

## GNS.

 Carr. 5/6 A complte kit of paris to construct a good quality $3+3$ watt (total 6 watt) stero amplifier providingreally life-like reproduction. Sultable for uase with all really life-like reproduction. Sultable for uge with all
steren pick-up heads at present avallable. Ganged volume and tone controls. Preset balance control Outputs for matched 2-3 ohm speakers. For 200-250 v. A.C. mains. Astonishing value.

## R.s.C. STEREO/TEN HIGH QUALITY

 AMPLIFIER KIT

POCKET PORTABLE TRANSISTOR RADIO DIESIGN. Employing 2 Brimar R.F. Transistors 1 output Transistor, and crystal diode, Ferrite Rod Aerial Minlature spenker unit. Handsome Plastic Case. Con

PRACTICAL WIRELESS SUPER SIX
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6 Transistor Superhet Radio. Full constructional details etc., $1 / 6$. All required parts including attractive plastic
case and dlal, printed circuit and first grade transistors. Only $£ 8 / 19 / 6$
EXTENSION SPEAKERS. Fiandsome walnut veneered


## SELENIUM RECTIFIERS

H.T. Types
120
H.W.
20

| L.T. Types | H.T. Types H.W. |
| :---: | :---: |
| ${ }^{2 / 6}$ \%. ${ }^{\text {f }}$ n.H.W. .- $1 / 9$ | 120 จ. $40 \mathrm{~mA} \ldots . .3 / 9$ |
| e/12 v. 1 a.H.W... $2 / 9$ | 250 v. 50 mA .... $3 / 11$ |
| Following F.W. (Bridge) | 250 จ. 60 mA |
| 6/12 ₹ \% 1 n. .l.... $3 / 11$ | ${ }^{250}$ v. 80 mA c. $6 / 11$ |
|  | 250 \% . 250 max ....12/9 |
| 6/12 v. 4 a. ......12/3 | 250 v. $80 \mathrm{~mA} \ldots .6 / 11$ |
| 8/12 จ. 5 n. ......14/8 | 250 จ. 50 mA F.W. |
| 12 v . 6 a | (Bridge) |
| $6 / 12$ จ. 10 a | 50 v. 75 mA F.W |
| 8/12 จ. 15 a . . . . 35/8 | (Bridge) $\quad \ldots . .10 / 11$ |
| RECORDING HEADS. Bai (boused in one contalner), | Record Playback and Erase 6 palr. |

## 1911

RECORDING HEAD8. Baird Record
(housed in one container), $9 / 6$ pair.
designed for simplicity in wriring. Sensitivity and quality are well up to standard. Point-to point wirng diagram for is maximum of $£ 4 / 19 / 6$ including reaber can be buill in brown or cream bukelite or venecred walnut.


JASON F.M. TUNER. Type FMT1. All parts including Dial, Punched Chassis and Valves. Power supply required 180 v .25 mA and 6.3 v .

EX GOVT.

## SMOOTHING CHOKES

$\begin{array}{lll}60 \mathrm{~mA} & 10 \mathrm{~h} .400 \text { ohms } & 3 / 11 \\ 80 \mathrm{~mA} & 20 \mathrm{~h} . \\ 900 & \text { ohms } & 5 / 11\end{array}$ 100 mA 5 h . 100 ohms $3 / 11$ | 150 | mA | 10 | h. |
| :--- | :--- | :--- | :--- |
| 100 | 10 mma |  |  |
| 10 |  |  |  | 120 mA 12 h 100 ohms $10 / 11 / 9$ $200 \mathrm{~mA} \mathrm{5-10h} 100$ obms $11 / 9$

## Battery

## HEAVY DUTY KIT

 8/12 v. variable charge rate up to ${ }^{6}$ amps. Trans., F.W. (Bridge) Selenium Rectifier, 0.7 Charge Selector. Fuses fuse-holders, panels, plugs and circuit. Only 59/6. Post $4 / 6$.CHARGER
TRANSFORMER
200-230-250 v. $50 \mathrm{c} / \mathrm{s}$.
$\begin{array}{llll}200-230-250 & \text { v. } & 50 & \mathrm{c} / \mathrm{s} . \\ 0-9-15 & \text { v. } 1 \frac{1}{4} & \mathrm{a} . & \cdots \\ 0 & 12 / 9\end{array}$

$\begin{array}{lll}0-9-15 & \text { v. } 3 \text { a. } & \text { … } \\ 0-9 / 15 & \text { v. } 5 \text { a. } \\ 0 & & 19 / 9\end{array}$
$0-9-15$ v. 5 a.
$0-9-15$ v. 6 a.
hargers and Kits for 200-230-250 v, $50 \mathrm{c} / \mathrm{s}$. A/C. Mains Consisting of Mains Transformer Consisting of Mains Transformer,
F.W. Bridge, Metal Rectifier, F.W. Bridge, Metal Rectifer, weil ventilated steel case. Fuses,
fuse-holders, grommets, panels fuse-holders, grommets, panels
and circuit.
Carr. $2 / 9$ extra. 6 v . or 12 v .1 amp .

## As above, with ammeter

## 6 v .2 amps.

6 v . or 12 v .2 amps.
6 v. or 12 v. 2 amps.
(inclusive of ammeter) 6/12 v. 4 amps.
6 v . or 12 v .4 amps., with variable charge rate selector

ASSEMBLED CHARGER
6 v . or $12 \mathrm{v}$.2 amps . Fitted Ammeter and selector plug for 6 v . or 12 v . Louvred metal case, finished attractive hammer blue. Ready for use with mains and output leads. Double Fused. Only
59/9
$50 \mathrm{c} / \mathrm{s} . \mathrm{A} / \mathrm{C}$. Mains
ASSEMBLED 6 v . or 12 v .


Fitted Ammeter and variable charge selector. Also selector plug for 6 v . fused. Well ventilated steel case with blue hammer finish. Ready for use with 69/9
mains and output leads. Carr, $5 /$ -
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$13 / 3$ and 5 monthly payments of $13 / 3$.
As above, but for 6 amp . charging 4 GNS. Carr. 5/-. Or Deposit 16/- and 5 monthly payments of 167 -

## . 50 micro-amp MICRO-AMMETERS

Flush mounting, 29/6.

## EX GOVT. MAINS TRANSFORMERS

Primary $0-110 \cdot 200-230-250$ ₹. 275-0-275 $7.100 \mathrm{~mA}, 6.3$
$\begin{array}{lll}\text { Input } 200-250 ~ 7 . ~ & 50 \text { c.p.s. } 250 \text { v. } 60 \mathrm{miA} \\ 6.3 & \mathrm{v} .2 \text { a... } 10 / 11\end{array}$
Primary $200 \cdot 250$ v. Sec. 12 v. 20 a .
Primary $200-240$ y gec $3500 \times 5 \mathrm{~mA}$
50 waite, $0.110 / 120$-230/250
D.C. sUPPLY KTTS, sultable for electric trains. Consist of mains trans. 200-250 r. 50 c.p.s. 12 r. 1 mpp. sele nlum
rect. (F.W. Bridge); 2 fugeholders, 2 fuse, change direction rect. (F.W. Bridge); 2 fugeholders, 2 fuses, change direction
switch, variable speed regulator, partinury drilled steel case and circuit. Very limited number, $33 / 9$

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SELENIUM RECTIFIERS
With large aquare aluminium cooling fins.
16 amp. F .W. (Bridge). Limited number.
15 amp .

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Well pentilated, black crackle finshed, undrilled caver Slze $14 \times 10 \times 8 \geqslant \mathrm{in}$. high. IDEAL FOR BATTERY CHAR GER OR INSTRUMENT CABE. COVER COULD BE USED FOR AMPLIFIER, Only 9/9, plus $2 / 9$ post.
ReLays. Carpenter Type Polarised, $2 \times 9,600$ turns at 1,685 ohms, 13/9. Miniature type G.E.C. 670 M1092

## R.S.C. A10 ULTRA LINEAR 30 WATT AMPLIFIER


gram, etc., can be simultanconsly applied for mixing purposes. AN OUTPUT SOCKET RADIO FEEDER UNIT, Price in kit form with eary to follow wiring diagrams, ONLY
Carr. $10 /$ - 11 OnS Or Factory built using latest ELB4 output valves and with 12
GNS Carr. 107. 11 GIS. months guarantee. 14 GEPS. $33 / 3$ and 9 monthly payments of 33/3. UNITS. Protertive Cover 19/9. Type 807/output ralves are used with High Quality sectionally feedhak of 20 D.B. in main loop. CERTIFIED PERFORMANCE FIGURES ARE EQUAL TO M at $12,000 \mathrm{c} / \mathrm{cs}$. hum and nolse $70 \mathrm{D} . \mathrm{B}$. down. Good quality reliable components used. Chassis finish blue hammer. Overall size $12 \times 9 \times 9$ in. approx. Power consumption 150 watts. For A.C. mains $200-230$ v. 50 c/s. Outputs for 3 and 15 ohm speakers. EQUALLY
SUTTABLE FOR THE CONNOISSEUR OR FOR LARGE HALLS, CLUBS OR OUTSIDE
FUNCTIONS, TDEAL FOR USE WITH MUSICALINSTRUMENTS, SUCH AS STRING BASS, FUNCTIONS, IDEAL FOR USE WITH MUSICAL INSTRUMENTS, SUCH AS STRING BASS,
ELECTRONC ORGAN, GUITAR, etc. FOR DANCE BANDS, GARRISON THRATRES, ete., eto. We can supply Microphones, Speakers, etc., at keen cash prices or on terms with ete, eto. We can supply Microphones, Speaker
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FULL RANGE OF LINEAR HIGH FIDELITY AMPLIFIERS ALWAYS IN STOCK, For $200-250$ GL, 50 c.p. A. A.C. mains. Overall size only $11 \% \times 2\} \times 2 \frac{1}{2} \mathrm{in}$. Fitted Vol. and Tone Control with malns switch. Designed for use with any kind of single player or record changer unit. Output for $2-3$ ohm speaker. Guaranteed 12 months. Only 59/6.

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A highly senaitive 4 -valve quality amplifer for the home, small club, etc. Only 50 millivolts input is required for full output so that it is suitable for use with the latest higb fidelity plek-up heads in axldition to all other types of pick-ups and practically all makes. equalisation. Eum-level is nerligible being 71 D.B. equw. $15 \mathrm{D} . \mathrm{B}$. of negative feedhack is used. H.T.
of 300 F .26 mA . and L. T. of 6.3 v .1 .6 . a. is a vailable for the supply of a Radlo Feeder Unit or Tapc Deck pre-amplifier. For A.C. malns input of $200-250$ v. $50 \mathrm{c} / \mathrm{s}$. Output for $2-3$ ohm speaker. Chassis is not fully punched chassis (with baseplate) with the blue hammer finish and polnt-to-point wiring diagrams and instructions. Exceptional value at only $£ 4 / 15 /-$
 or assembled ready for use $25 /-$ extra, plus $3 / 6$ car- $2 /$ - for assembled unit.
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12 in. R.A. $29 / 11$. 12 in. B.A. 3 or 15 ohms, 10 watts, 12,000 lines, $59 / 6$.

## TWEETERS, 4in. Plessy, 3 ohms 18/9. E.A. 15 ohms $25 / 9$.

R.S.C. TRANSFORMERS Fully Guaranteed MAINS TRANSFORMERS. Primaries $200-250-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$. PULLY SHROUDED. UPRIGHT MOUNTING.
$250-0-250$ v. $60 \mathrm{~mA} ., 6.3$ v. 2 A., 5 ₹. 2 a. $2 \frac{1}{2}-3-3 \mathrm{~m}$.
$250-0.250$ v. 100 mA .6 .3 v. 4 a., 5 v. 3 a.
 $350-0-350$ v. $100 \mathrm{~mA} ., 6.3$ v. 4 a., 5 v. 3 a.

 TOP SHROUDED DROP-TEROUGH TYPE | $260-0-260$ |
| :--- |
| $250-0-250$ |
| v. |
| v |
| $0 \mathrm{~mA}, 60 \mathrm{~mA}$ |
| 10.3 | $250-0-250 \mathrm{\nabla} .100 \mathrm{~mA}$

$250-0-250$
v .100 mA

| $250.0-250$ |
| :--- |
| $350-0-350$ |
| v. |
| v |
| 100 mA |

$350-0-350$ v. $80 \mathrm{~mA} . .6 .3$ v. $2 \mathrm{s.,5} 5 \mathrm{v} .2 \mathrm{a}$.
$300-0.300$ v. $100 \mathrm{~mA} ., 6.3$ マ. 4 ฉ., 5 v. 3 マ. 4 а., 5 v. 3
$300-0-300 \mathrm{v} .130 \mathrm{mAA}, 6.3 \mathrm{~V} .4$ a., c.t., 6.3 v .1 a

suitable for Mullard 510 Amplifler | suitable for Mullard 510 Amplifler. |
| :--- |
| $350-0-350$ |
| v. | 00 mA . 6.3 v. 4 a


$350-0-350$ v. 150 mA .6 .3 v .4 a., 5 v. 3 a
$425-0.425$ FILAMENT TRANSFORMERS


## AUTO (Sten Up/Step Down) TRANSFORWERS

 $50-80$ watte $110-120 \quad \nabla . / 230-250$1.50 watts $110-120 \quad$ v. $200-250 \quad \nabla$.

Interleaved \& impregnated. OUTPUT TRANSFORMERS Midget Battery Pentode $\begin{array}{ll}66: 1 & \text { for } 384, \text { ete... } 3 / 8 \\ \text { Small } \\ \text { Pentode } \\ 5,000\end{array}$ to $3 \Omega$ Pentode $5,000 \Omega$ Standard Pentode 5,0000 Standard Pentode $8,000 \Omega$ to 3 Multi Ratio, Single or P/P 3 or 150 Push-pull 8 watts EL84s to 3 ohms
Push-pull 8 watts EL84s to 15 ohm Push-pull $10-12$ watts 6 V , to $3 \Omega$ or $15 \Omega$

LINEAR L45 MINIATURE 4/5 W. QUALITY AMPLIFIER. Suitable for use with any record playing unlt and most micropho nes. Negative feedback $12 \mathrm{D} . \mathrm{B}$, bass and treble
controls. For A.C. mains input of $200-250 \mathrm{v} .50 \mathrm{c.p.s}$. Output for $2 / 3$ ohm speaker. Three milniature Mullard valves.
Blze only $6 \times 5 \times 511 \mathrm{n}$. high. Chassis fully isolated from
 meposit $22 /-$ and 5 monthly payments $\mathbf{2 5 . 1 9 . 6}$ of $22 /=.$. Send $\operatorname{S.A}$.E. for leaflet.

## HIGH FIDELITY 12-14 WATT AMPLIFIER TYPE A11



Two lnput sockets with associa-
of " mike" and gram. as in
ECC83,
EL84, EL84, 5Y3. HIgh Quality sectionally wound output transformer specially designed for Ultra Linear operatlon and reliabie small condensers of current manufacture. IN
DIVIDUAL CONTROLS FOR BASS AND TREBLE " Lift " and Cut." Frequency DIVIDUAL CONTRO18 FOR BASS AND TREBLE "Lift" and "Cut." Frequency reaponse $\pm 3$. B. $30-30,000$ c/cs. $81 \times$ negative feedback loops. Hum level 60 D.B. down makes and types of pick-ups and microphones. Comparable with the very best designs For STANDARD or LONG PLAYING RECORDS. For MUSICAL INSTRUMENTS such as TRING BASS, GUITARS, etc. OUTPUT SOCKET wlth plug provides 300 v .39 mA . and $6.3 v .1 .5$ a. For supply of a RADIO FEEDER UNIT, Size approx. $12 \times 9 \times 7 \mathrm{in}$. For A.C mains $200-250$ \%. 60 o/cs. Output for 3 and 15 ohm speakcrs. Kit is complete to last nut Chasses is fully punched. Full instructions and point-to-point wirtng 8 GnS. Carr
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with cash and credit terms.

## R.S.C. PORTABLE GUITAR AMPLIFIERS



## JUNIOA 5 WATT. High Quality Output. Separate Bass and Treble "cut" and "a boost " senarate

 Sensitivity 15 mv . Bigh Flux 8 in . $1 /$ speaker. InputBockete for Radio/Tape or Gram Pick-up and Mike Bockets for Radio/Tape or Gram Pick-up and Mike Instrument Pick-up. Handsome atrongly made attractive and durable polychrome and fitted carry C8.19.6 Carr. 7/6. Or Deposit $£ 1$ and nine Send S.A.E. for leafiet payments \&1.
SENIOR 10 WATTS. High-Fidelity Push-Pulloutput. Separate Bass and Treble "cut" and high gain inputs so that iwo instruments such as Guitar and String Kass can be used at the same time. Two Loudspeakers are incorporated in 12in. P.M. for Biss notes and $17 \times 41 \mathrm{~m}$. elliptical for Iunior model. Bize approx, $18 \times 18 \times 8 i n$. 15 Gns. Plus $10 /$ carr. H.P. TERMS DEPOSIT $34 / 9$ and 9 monthly payments $34 / 9$. Both models for $200-250$ r. A.C. mains.
COLLARO OONQUEST 4-SPEED AUTO-CHANGERS. With studio pick-up with turnover head. lateat model for 200-250 v. A.C. mains, $26: 18 / 6$ B.S.R. MONARCE AUTO-CHANGERS. Type UAs. 4 speed T/O Plck-up with sapphire stylus £6/19/6. Carr. $4 / 6$.
Any of the above supplied with T/O stereo/monaural head for f1 extra.
COLLARO JUNIOR. 4 -speed Bingle Players with Hi-Fi T/O crystal pick-up bead, $£ 3 / 19$
LOUDSPEAKER IN POLISHED WALNUT FINISHED CABINET- Gauss 12,000 lines Speech coil, 3 ohms or 15 ohms. Only e4/19/6.
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12in. 20 WATT 15,000 line $1 /$ speakers 15 ohms in Cablnet finisked as above. Size $18 \times$ $18 \times 8 \mathrm{in}$. $\mathrm{E} / 19 / 6$ or Deposit $17 / 9$ and 9 monthly payments of $17 / 9$.

LINEAR TAPE PRE-AMPLIFIER Type LP/1. switchen negative feedback equalisation. Positions for Recorl
17in.. 311n.. 7hin, and Playback. EM84 Recorilnr level indicator. Designerl primarily as the link letween Collaro Tape Transeriptor and higb fidelty amplifier but suitable almost any Tape Deck. 9 GNS. Scnd \$.A.E. for leaflet.
R.8.C. STANDARD BASS REFLEX CABINET for 12in Loudapeakers, Acoustically lined and ported. Size 20in. $14 \mathrm{in}, \times 13 i \mathrm{v}$. Beautifu wainut veneer inish. Especialy recommended for use with speaker below \&5/19/6.
get of four legs can be supplied for $29 / 6$ per set.

## PLESSEY

 CONCENTRIC12in. P.M. SPEAKERS
(15 ohma) consisting of a
high quality 12in. speaker of orthodox uesign support Ing a small elliptleal speakand condensers to act as tweeter. This high adelity unit is highly recommended for nse with our All or any
gimilar amplifier. Eating Js 10 watts. Gauss 12,000
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Or Deposit $13 / 9$ and 9 Or Depost $13 / 9$ and 9
monthly payments of $13 / 9$.

# CWINTH <br> 追 <br> (BADIO) LINITED <br> Phone: GERRARD 8204/9155 Cables: SMITHEX LESQUARE LISLE STREET, LONDON, W.C. 2 



## R.C.A. AR88D RECEIVERS

This world famous 14 valve receiver offered reconditioned, perfect working order and in superlative condition throughout. Frequency coverage on 6 bands $500 \mathrm{kc} / \mathrm{s}$ to 32 $\mathrm{mc} / \mathrm{s}$. Circuit incorporates variable selectivity with crystal filter, tone control, aerial trimmer, b.f.o., a.v.c., R.F. and A.F. gain controls,
mechanical bandspread, etc. Output is for mechanical bandspread, etc. Output is for
phones or speaker. Operation 115 or 230 phones or speaker. Operation 115 or 230 volts A.C. $£ 35$ each. Carriage $30 /$.
BRAND NEW MEDRESCO HEARING AIDS
Supplied complete with earpiece, leads, battery pouch. Brand new boxed, only 32/6 each P/P I/-. Batteries 5/- extra.

## COLLARO STUDIO <br> TAPE TRANSCRIPTORS

Latest 1961 model, 3 speeds, 17 , $3 \frac{3}{4}$ and $7 \frac{1}{2} \mathrm{in} . / \mathrm{sec}$. Fitted with 3 separate motors, digital counter press burton switching, provision for fitting extra stereo head. Fitted with new Bradmatic heads. Supplied brand new with instructions, complete with spare 7in. spool. $£ 12$ each. P/P $3 / 6$.

AVO SIGNAL GENERATORS
Frequency coverage $95 \mathrm{kc} / \mathrm{s}$ to $40 \mathrm{mc} / \mathrm{s}$. Ideal for all general radio work Variable attenuator, 400 cycle int. mod. or provision for ext. mod. Supplied fully tested and checked $£ 7 / 19 / 6$ each. P/P $3 / 6$. Operation is from 2 v , and 60 v . batteries, but original Avo mains units can be supplied at $19 / 6$ extra.
MINE DETECTOR NO. 4A
Will detect ferrous and non-ferrous metals. Complete and as new in transit cases. Supplied fully tested with instructions. $39 / 6$ each. Carriage 10/-. Batteries $8 /$ - extra.

## MARCONI TF-373 UNIVERSAL

IMPEDANCE BRIDGES
Reconditioned to maker's specification. $0-100$ Reconditioned. to maker's specification. 0-100
$H ., 0-100 \mathrm{mfd}$., $0-1$ megohm, 0-100 Q. each on 5 ranges at $1,000 \mathrm{c} / \mathrm{s}$. $€ 35$ each.

VALVE VOLTMETERS No. 2 Laboratory instruments. Five ranges A.C. and D.C. 1.5 v. 5 v. 15 v., 50 v. and 150 v. Operation $200 / 250$ volts A.C. Supplied as new, fully tested and complete with internally mounted H.F. probe. $\mathrm{f} 17 / 10 / \mathrm{e}$ each. Carriage $10 /$ -

## R1155 RECEIVERS MODEL L/N

This model incorporates the $100-200$ metre top band. Full coverage on 5 bands is $200 \mathrm{kc} / \mathrm{s}$ to $18 \mathrm{mc} / \mathrm{s}$. Supplied fully tested and in perfect condition $£ 12 / 19 / 6$ each. Carriage $7 / 6$. A combined $200 / 250$ volt A.C. power unit and audio output stage to match 3 ohms supplied extra at $85 /$-.

## AMERICAN ARB RECEIVERS

Frequency coverage on 4 bands $195 \mathrm{kc} / \mathrm{s}$ to $9.05 \mathrm{mc} / \mathrm{s}$. Precision vernier drive. Valve lineup: 12SA7, 4-12SF7, 12A6 and 991. Operation 24 volts D.C. Supplied fully tested and checked, 46/19/6 each. Carriage 7/6.
MULTI-RANGE TESTMETERS MODEL
1,000 ohms per volt A.C.ID.C. Ranges: Volts A.C. $10,50,250,500,1,000 \mathrm{v}$. Volts D.C. 10, 50, 250, 500, 1,000 v. Current D.C. I mA 100 mA . 500 mA . Resistance, 2,000 ohms, 200,000 ohms. Supplied brand new and guaranteed complete with instructions and leads 59/6 each. P/P 2/6.

SPARES KITS FOR CR. 100 RECEIVERS. Contains 15 valves. 2-DH63, 2-KT63, 2-X66, 2-U50, 7-KTW61. Condenser and resistor packs, pots, toggle switch, output transformer, etc. All brand new, $59 / 6$ each. P/P 3/6.

## NATIONAL H.R.O. RECEIVERS



Senior model, table mounting. Supplied with complete set of 9 coils covering 50 Circult incorporates Circuit incorporates selectivity, erystal filter, 2 R.F. and 2 I.F. stages, B.F.O. A.V.C., R.F. and A.F. gain controls, etc. Output is for phone or speaker. Power requirements 250 v. 80 mA . and 6.3.v. 3 amps. Supplied fully tested and aligned and in superb condition throughout. Price 21 gns. Carriage $10 /$-. Power units are available $59 / 6$ exera.

## AR88D SPARES

Complete wavechange switch assembly with screens. New boxed, $17 / 6$ ea. P/P 2/6. Ist 1.F transformers. New, boxed, 3/6 each. P/P 9d.

PAINTON MINIATURE JONES PLUGS
AND SOCKETS

| All new and unused. |  |  |  |
| :---: | :---: | :---: | :---: |
| 2 pin | ........... 2/6 pr. | 12 pin | $5 / 6 \mathrm{pr}$. |
| 4 pin | ........... 3/6 pr. | 18 pin | ...... $7 / 6 \mathrm{pr}$. |
| 6 pin | 4/-pr. | 24 pin | . 816 |
| 8 pin | 4/6 pr. | 33 pin | $10 / 6$ p |

## PARMEKO TABLE TOP TRANSFORMERS

Input 230 voles 50 cycles. Output 620/550/ $375 / 0 / 375 / 550 / 620$ volts 250 mA . 5 volt $3 \mathrm{amp}, 5$ volt 3 amp . Size $6 \frac{3}{4} \times 6 \frac{1}{4} \times 5 \frac{1}{2} \mathrm{in}$. Brand new boxed. 45/- each. Carriage $3 / 6$.

ROTARY CONVERT゙ORS
24 volt D.C. Input. Output 230 volts 50 eycles 100 watts, $72 / 6$ each. Carriage 5/m.
R.C.A. PLATETRANSFORMERS

Primary 200/250 volt 50 cycles. Secondary $2,000 / 1,500 / 1,500 / 2,000$ voles 500 mA . Supplied brand new boxed, $£ 6 / 10 /$ each. Carriage $10 /$.

POST OFFICE 8 BANK UNISELECTORS Std, type, 25 position. Ex-brand new equipment. 62/6 each. P/P 2/-.

## AN/APR4 SEARCH RECEIVERS

Covers 38 to $1,000 \mathrm{mc} / \mathrm{s}$ with 3 plug in R.F. units. TN 16 Covers $38-95 \mathrm{mc} / \mathrm{s}$. TN $1774-320 \mathrm{mc} / \mathrm{s}$. units. TN $16,38-95 \mathrm{mc} / \mathrm{s}$. TN $1774-320 \mathrm{mc} / \mathrm{s}$.
TN18 $300-1,000 \mathrm{mc} / \mathrm{s}$. Operation 115 v . $50-2,600 \mathrm{cps}$. Reconditioned to maker's spec. Superb order throughout, $\notin 75$ each.

## SANGAMO WESTON STANDARD

 VOLTMETERSRange $0-30$ voles D.C. 1,000 ohms per volt. Correct to B.S. 89 pr. limits. 6in, mirror scale, $\epsilon 25$ each.

SUB-STANDARD I.F. ALIGNMENT OSCILLATORS. 3 ranges covering 445 to $485 \mathrm{kc} / \mathrm{s}$. Crystal controlled and fitted with precision variable attenuator. Operation 200/250 volts A.C. Brand new, $\& 15$ each Carriage $10 /$ -

PLESSEY
24-VOLT
D.C.

## PUMPS



Self lubricating, capacity 60 g.p.h. at 30 lb . $\begin{array}{lllll}\text { sq.ec/outlet union. Only } 15 / 6 \text { each. P/P } & 2 / 6 .\end{array}$



## TAPE AWPLIFIERS and PREAMPLIFIERS presented from MULLARD DESIGNS

ERASE UNIT

MULLARD TYPE "C" TAPE-PREAMPLIFIER

The " Ei-Fl'" link to add full tape recording facluties to High Fidelity
home ingtallations. Incorporateg
FERROXOUBE POT CORE PUSE PULL OSCLLLATOR and 3 -speed treble equalisation by FERROXCUBE POT CORE TNDUCTOR. FOR WEARITE-COLLARO TBUVOX or BRENELL TAPE DECKS. (BTATE which when ordering.) Includes Beparate Power Supply Uaito \$14.0.0 \$17.0.0



MODEL HF/TR3 Mk. II TAPE AMPLIFIER

## (Mullard Type "A" design)

 A very high quality Amplifier Incorpora-ting 3 -speed
treble equalisation, by the latest FERBOXCUBE POT CORE
INDUCTOR. FOR COLLARO-TRUVOX INDUCTOR FOR COLLARO-TRUVOXBRENELL of WEARITE Tape Decks (STATE which when ordering), has
GILSON Output Trauslormer, Includes separate Power Supply Unit.
KIT OF PARTS .......
13.13.0
 G.P. E3181- Depor ASSEMBLED E.P. £318/- Deposit and 12 mont

## FOR THE HOME CONSTRUCTOR SPECIAL "COMBINED ORDER" PRICES

(o) The COLLARO "STUDIO" TAPE DECK and our Mullard Type "C"PRE-AMPLIFIER and Power Unit assembled and rested H.P. Terms: Deposit $65 / 18 /-$ and 12 months at $£ 2 / 3 / 3$.
(b) As above but Type "C" PRE-AMPLIFIER supplied as
(c) The TRUVOX Mk. VITAPE DECK and the assembled 840.0 .0 H.P. Deposit $£ 8$ and 12 months $£ 2 / 18 / 8$.
(d) As above but the Type " C " supplied as complete $£ \mathbf{£ 3 6 . 1 0 . 0}$
(e) The BRENELL Mk. $V$ Deck and the asssmbled Type $£ 46.0 .0$ H.P. Deposit $£ 9 / 4 /-$ and 12 months at $£ 3 / 7 / 6$.
(f) As above, but the Type " C " supplied as complete

The WEARITE 4A DECK with Type " $C$ " assembled and tested $H$. . Deposit $\mathbf{\epsilon} 1 \mathbf{1} / 4 /-$ and 12 monthly $\epsilon 4 / 2 / 1$.
H.P. Deposit $£$

EACH OF ABOVE CAN BE SUPPLIED IN PORTABLE CASE FOR 65/10/-
extra. THUS FORMING A COMPLETE PORTABLE PRE-AMPLIFIER.
SEND FOR DETAILS. $\quad 225 \mathrm{ft}$. on 3 in . Spool ....... $5 / 9$
 $\begin{array}{llll}\text { P.V.C. base on latest type plastle } & 1,20 \mathrm{ft} \text {. on } 5^{\frac{3}{3}} \mathrm{in} \text {. Spool ..... } & 21 /- \\ \text { Spools. New, Boxed and Guar- } & 1,200 \mathrm{ft} \text {. on } 7 \mathrm{in} \text {. Spool ...... } & 21 /-\end{array}$ i,800ft, on 7in. Spool …… $32 / 6$
(a) COMPLETE KIT to build the HF/TR3 Amplifier, together with the COLLARO "STUDIO" DECK (b) As above, but HF/TR3 ASSEMBLED and TESTED H.P. Terms: Deposit $£ 5 / 18 / \mathrm{-}, 12$ months of $£ 2 / 3 / 3$ (c) COMPLETE K (c) COMPLE NEW TRUVOX Mk. VI TAPE DECK ................ $\$ 36.10 .0$
(d) As above but HFITR3 ASSEMBLED and TESTED H.P. Terms: Deposit $£ 8,12$ months of $£ 2 / 18 / 8$.
(e) COMPLETE KIT to build the HF/TR3 AMPLIFIER
the BRENELL. Mk. V TAPE DECK .................................
(f) As above but HFITR3 ASSEMBLED and TESTED H.P. Terms: Deposit $£ 9 / 2 /-, 12$ months of $£ 3 / 6 / 9$.
(g) THE ASSEMBLED and TESTED HF/TR3 AMPLIFIER with the WEARITE MODEL 4A DECK, Incorporates Wearite Head Lift Transformer, etc. $\%$. $1 / \ldots \%$.
H.P. Terms: Deposit $£ 11,12$ months of $£ 40 / 8$.
(Carriage and insurance on each above is $10 /$ - extra.)
Attractive PORTABLE CASE is available to accommodate the TRUVOX Or COSTION $10 \times 6 \mathrm{in}$. LOUDSPEAKER-ACOS CRYSTAL MICROPHONE -and $1,200 \mathrm{ft}$. SPOOL. TAPE-ALL FOR..

| TAPE ACCESSORY KITS |
| :--- |
| (a) E.M.I., includes 3 reels Leader Tape, Splicer, Jointing Tape |
| and Stop Foil $\ldots . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . ~$ |

(b) SCOTCH BOY̌, Includes 3 reels Leader Tape, Splicer, and $29 / 6$

FULLY GUARANTEED TRUVOX and GARRARD TAPE EQUIPMENT ENABLES THESE OUTSTANDING PRICE REDUCTIONS THE "MODEL HF/G2R" PORTABLE TAPE RECORDER (Original Price £33.0.0)
 INCORPORATES THE LATEST GARRARD "MAGAZINE" TAPE DECK and MATCHING AMPLIFLGR. Based on the saccessful specifically developed to operate the GARGARRARD TAPE MAGAZINE and 4in. BPOOL OF DOUBLE PLAY TAPE. A sec. providing up to 1 hour 10 ming. playing time. The aimplicity of operation. Incorporates EXT. SPEAKER SOCKET, from P.U., mike or Radio tuner. Welghs oniy 22 bl . WE ALSO OFFER DECE and AMPLIFIER OONNECTED, TESTED, FOR IMMEDIATE OPRRATION, 19 gns. H.P. Dep. £4 and 12 months £1/9/4, Carriage and Ins. 101 ex. a complete tape reconder chassig ready for easy ftting into ce blinet.

THE "MODEL TK/MK. IV " PORTABLE TAPE RECORDER (Original Price £49/10/-)
 ONLY Dep. 87/6/- and 12 months EMI TAPE. NCORPORATES THE TRUVOX Mk IV TAPE DECK, ROLACELESTION 98 Sin, LOUDFIER specifically developed by Truvox Lidd. to correctly operate their MSk IV Tape Deck.
Thls combination affords first-class tape re. This combination
A Twin Track Two Speed model operating at
84 and 7 inin. sec . Incorporates SAFETY BUT34 and 7 lin. /sec. Incorporates SAFETY BUT-
TON (prevents accide ntal erasure). Ext. Speaker. TON (prevents accidental erasure). Ext. Speaker.
TONE And VOLUME CONTROLS. Also operates
 as independent AMPLIFIER for direct
tion from P.U., mike or Radio tuner.
WE ALSO OFFER THE DECK and AMPLIFIER AS FOLLOWS: Ms. IV TAPE DECK. £18/10/-. H.P. Deposit $£ 3 / 6 /-12$ months $£ 1 / 4 / 3$. TYPE "K" AMPMFIER, 1215.
H.P. Deposit $£ 3.12$ month $121 / 9$. COMBINED ORDER FOB BOTE DECK AND H.P. Deposit $£ 3$.
AMPLIFIER,
$£ 30$.
H.P. Deposit
I

# STERN'S MULLARD DESIGIS 

 OF PARTS

D-signed by MULLARD-presented by MULLARD " 5-10" MAIN AMPLIFIER For use with the MULLARD 2 -gtage pre-amplifier with which an undistorted
power output of up to 10 wats is obtalned. We supply SPECIFIED COMpower output of up to 10 wats is obtalned. Wo supply sPECIFIED COM. TRANSFORMER and choice of the latest Ultra-linear PARMEKO MAINE PARTRIDGE Output Transformer.
Price: COMPLETE KIT (Parmeko O/put Trans.) ........... $8 \mathbf{\& 1 0 . 0 . 0}$
Alternatively we supply ASsembled and Tested ...... £11.10.0
ABOVE INCORPORATEG PARTRIDGE OUTPUT TRANSFORMER £1/6/-extra

## MULLARD'S 2-VALVE

PRE-AMPLIFIER TONE CONTROL UNIT amploying two EP86 ralves and designed to operate uplth the Mullaril MAIN AMPLIFIER but also perfectly Buitable for other makes. Equalisation for the latest R.I. A.A. characteristices.
lapuu for crystal Pick-ups and variahle retuctance magnetic types

- Sensitive Microphone Channel. Wide range BAss (band a Tape Amplifier or Pre-Amplife

Price: COMPLETE KTY $\underset{\text { OF PARTS }}{ } \mathbf{6} 6.6 .0$


## COMPLETE MULLARD 5-10 AMPLIFIER

 The popular and very guccessful complete " 5 -10 " incorporating Control Unit providing up to 10 watts high quality reproduction. Specifled components and new MULLARD VALVES are supplied including or Partridge Ultra Linear Output Transformers Price: COMPLETE KIT. Parmeko Trineformer.$£ 11.10 .0$ Alternatively we supply ABsEMBLED AND TEGTED. £13.10.0 Hire Purchase (Assembled Amp. only). Deposit £2/14j-, 12 months at $19 / 10$ ABOVE incorporating PARTRIDGE OUTPUT TRANBFORMER $£ 1 / 6 /$-cxtra


COMPLETE MULLARD 3-3 A VERY HIGH QUALITY AMPLTFIER DEVEL OPED PROM THE VERY POPULAR 3-VALVE MULLARD LABORATORIES. Price for COMPLETE KIT
OF PARTS
$\mathbf{~} 7.10 .0$ (Plus $6 / 6$ carriage and insurance). Alternatively supplied Assembled AND FULLY TEETED (Plus
carriage and insurance). 16.
$\mathbf{8 8 . 1 9 . 6}$
H.P. TERMS: Depooit $£ 2$ and 8 monthly payments of $£ 1$ Our kit is complete to the MULLARD specifcation fnctuding supply of specifed components, valves and PARMEKO OUT PUT TRANSFORMER. We also
include suitched inputs for 78 and L.P. records plus a Radio position. Extra power to include suftched inputs for 78 and L.P. record
drive a Radlo Tuning Unit is alao avallable.

COMPLETE STEREO AMPLIFIER
Meets the many requests for a low priced but good quality stereophonic Ampliner. Out KIT OF PARTS.......... $£ 8.10 .0$
Mk. \|" "Fidelity" FM TUNING UNIT

An attractlvely presented Unit incorporating MULLARD PERMEABILITY TUNING
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FOR THE CONSTRUCTOR .... £10.0.0 or ASSEMBLED.... £14.15.0

## SPECIAL CASH ONLY OFFER!!

The very attractive PORTABLE AMPLIFIER CASE matched P.M. SPEAKER. ALL FOR ONLY
(Plus $7 / 6$ carr. and ins.). The Amplifier
consists of
2 sestage design ineorporating the 3 modern BYA valves and has separate BABS and TREBLLE CONTROLS. The The Portable Case will also accommodate aimost finished in Grey Colour Rexine-WE ALso BUPPLY SEPARATELY:-
a) The 2 -stage (plue Rectifiet) AMPLIFIER b) The Portable carrying case c) Gilin. P.M. SPEAKER. $\begin{array}{r}2317 \\ 18 \\ \hline\end{array}$

"Hi-Fi" LOUDSPEAKERS WE HAVEN STOCKA GOODMANS-WHARFEDALE-W.B. ILLUSTRATED AND PRICED LEAFLETS ON REQUEST


## PRICE REDUCTIONS

(a) The COMPLETE KIT OF PARTS to bulld
both the $" 5-10 \%$ Main Amplifier and the 2
both the " $5-10$ " Main Amplifier and the 2
(b) The " 5 -10 " and the 2-Stage Pre-Amplifer
oth Assembled and Tested.
£18.18.0
(c) The COMPLETE KIT OF PARTS to build the Dual Channel "3-3" Amplifier and the Dual Channal Pre-Amplifer Control Unit.
(d) The Dual Cbannel " 3.3 " Ampliter and the Dual Channel Pre-Amplifer Control Unit both H.P. TERMS: Desped
£21.10.0
H.P. TERMS: Deposit $£ 5$ and 12 months of $£ 1 / 16 / 8$.
(e) The complete kit of Parts to build
one " $5-10$ " Main Amplifer (Piarmeko Trans-
former) and the Dual Channel Pre Amplifer Control Unit
$\$ 21.10 .0$
(f) One " 5 -10" Amplifer (Parmeko Trans. former) and the Dual Channel Pre-Ampifier
both Assembled and Tested
$£ 25.0 .0$ H.P. TERMS: Deposit $£ 5$ and 12 months of $£ 1 / 16 / 8$.
(g) COMPLETE KIT OF PARTS to build Two "5-10" Main Amplifers (incorporating Channel Pre-Amplifer Control Ond the Dual (h) Two " 5-10 "Amplifers (Parmeko Output Transiormers) and the Dual Channel PreTested H.P. TERMS: Deposit fig/4-- and 12 months of $£ 2 / 12 /-$ Carriage and insurance $7 / 6$ extra.
Priees quoted are subject to $£ 1 / 6 /-$ extra for Partridge Trans

## MULLARD FOUR CHANNEL

MIXING UNIT

## STEREO DUAL CHANNEL PRE-AMPLIFIER

This model Incorporates two 2 -vilve Pre-Amplifiers
(described above) combined into a Slagle Unit enabling it to be used for both STEREOPHONIC and MONAURAL operation. It Is designed primarily to operate with our range of MULLARD MAIN Aith any make of Amplifiers requiting an well of $250 \mathrm{~m} / \mathrm{v}$.


Price : Complete
KIT 0 P PARTB
£12.10.0 Aternatively
and TESTED H.P. Terms on assembled unit: $£ 3$ Deposit and 12 months of $£ 1 / 2 /$.

## STEREO " 3-3" MAIN AMPLIFIER

Comprises two MOLLARD 3-3 Maln Amplifers on one chassig. Operates with above
MULLARD STEREO PRE-AMPLIFIER. Output power 6 Fatus. Inputs for Crystal Pitk-up and Radio Tuner
KIT OF PARTS............ $£ 10.0 .0$ or ASSEMBLED...... $£ 11.15 .0$
! ! RECORD PLAYERS! Many at REDUCED PRICES ! ! !
Senas.A.E. tor ILLUSTRATED LEAFLET
THE EMI 4-speed single record player 4 gns,
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mick-Up The NEWiARO Model RP594, 4-speed Single Record Player,


Studio Cartridge
THE E.M. $4-$ speed Single Record Player, incorporating a high output

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Both avilable incorpo
ap onating the B.S.B. STEREO Pick-up, plays
Both available incorporating the B.S.B. STEREO Pick-up, piays
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GARRARD RCOO 4 -speed Autochanger atted with latest Crystal
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Phek latest GARRARD TRANSCRIPTION MOTOR " $301 "$.
£6. 9.6
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A RANGE OF "EASY TO ASSEMBLE" PREFABRICATED CABINETS Designed by the W. B. "BTENTORIAN" COMPANY for "Hi-F " " Loudspeaker systems or to accommodate high quality equipment. The acousteally deaigned Bass Retilex
Cabinets contalning the very sucessful "stentorian" speakers give really first-class Cabinets contalning the rery successful stentrian apakers give really irst-class
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SMALL HIGHSPEED MOTORS


Robust, high-quality, fancooled motor built to aircraft standards by English Electric. Continuously rated for $11,000 \mathrm{r} . \mathrm{p} . \mathrm{m}$. from 115 volt 3 phase 400 cycle supply. Only $4 \frac{1}{2} \times 2$ inches dia. with ${ }_{\frac{3}{6}} \mathrm{in}$. dia. fibre gear pinned to $3 / 16 \mathrm{in}$. dia. shaft which protrudes $\frac{1}{2} \mathrm{in}$. from end face. Substantial terminal block.
Brand New 30/- each.
Post paid.

## Master and Remote

 CONTACTORSTHE MASTER CONTACTOR is a robust high-quality spring driven clock with a Services quality balanced escapement driving a low friction pair of contacts that " make" every half-second. The enclosed mechanism is optionally maintained at a constant temperature by a thermostat and small heating element of 12 to 24 volts is applied to external leads provided. These, together with the contact leads, are brought out through torroidal filter units incorporated in the metal base of the unit-the whole of which is fitted in a practically sound-proof, temperature and vibration eliminating sorbo-rubber-lined wooden box, approx. 6 inches cube. Winding key and stop/start knob accessible on removing snap-on lid. THE REMOTE CONTACTOR is a solenoid operated ratchet mechanism turning a pointer at one rev. per minute over a two
inch dial with adjustable zero and a divisions. The solenoid was designed to be energised by 24 volt pulses at half seconds via the designed to be energised by 24 volt pulses at half seconds via the master contactor and the ratchet wheel is secured to a fibre cam that
opens a pair of contacts for 3 of each rev, and lets them close for opens a pair of contacts for of each rev. and lits them clate.
$\begin{aligned} & \text { Infirst class condition, guaranteedfully } \\ & \text { serviceable } \\ & \text { Post paid. }\end{aligned} \quad 35 /$ a pair

## SUPPRESSOR SNIP

Neat, black, die-cast, four partition box, $3 \frac{1}{2} \times 4 \times 1 \frac{1}{2}$ in., with central screwed outlets. Two empty sections and two containing cascaded RF filters each comprising two $0.01 \mu \mathrm{~F}$ non-inductive 350 v . condensers and associated pi-wound coils clamped in die-cast cores mounted on easily removable ebonite blocks.
TYPE P. 2 BRAND NEW 3/6 each. Post paid.
TTP: Buy two, fit four filters into one box, steamroller that mains borne interference and have a useful spare box for building a screened pre-amp or oscillator unit into, all for $7 /=$. POST PAID

## 400 CYCLE CHOPPER

Latest version of U.S.A. Servomechanisms 400 c/s chopper. High quality $6 \quad v . \quad v i b r a t o r$ oscillating between twin contacts for chopping external circuit. Hermetically sealed in octal based can.

Brand
New
82 past
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Receiver
Type 1143A
Suitable for conversion to 2 metre FM or Wrotham Transmissions. Valves: 4xEF50, 1xEL 32 2xEF39, 1xEBC33, 1xEA50. Complete with circuit diagram, 81.5 .0 Plus
5/- packing and carriage.

## Cold Cathode Trigger Tubes

A sub-miniature cold cathode valve developed by Ericsson primarily for computer work, these GTR. 120W tubes have great possibilities in a number of experimental electronic automatic control circuits. They have an Anode-Cathode running voltage of 95 to 140 at 4.5 mA , and at 290 anode volts require a trigger current of only 250 microamps to cause the anode to take over the discharge. Typical ionization time $=90$ micro-seconds. They will withstand up to 310 v . with zero trigger voltage without self-igniting.
Supplied complete with full performance data
in original packs of 100 at the Special Price of
per 100 post paid

## MEGISTORS, 125, 1,000 or 10,000 MEGohms

Glass encapsulated $10 \%$ tolerance high value resistors for minute grid current applications. Ideal for extending the range of sensitive meters or using in probes to provide a really high impedance input for VTVM's or 'Scopes.
One of each value plus any chosen two, 5 for 10
post paid.

## $25 \mathrm{c} / \mathrm{s}$ Tuning Fork Drive AMPLIFIER

 Modern, light-alloy cased, Drive Unit, Type 114, containing a robust $8 \frac{1}{2}$ inch induction sustained 25 cycles per second tuning fork with attendant induction pick-ups and wave form amplifier comprising output. stabiliser in power supply derived from high-cycle transformer easily replaced with standard 250 v. mains type providing centre tapped H.T., 5 v. for rectifier and two 6.3 v . filament windings for valves (6L6 is heated from separate winding to EF50's and diode). High grade components throughout. Tuning fork assembly 3-point flexibly mounted, quickly removable, energised by 6.3 v . A.C. Overall size $8 \frac{1}{2} \times 7 \frac{1}{2} \times 10 \mathrm{hin}$.
£3.3.0 Carriage paid.

## CENTRE SCALE COUNTERS

Ex-RAF camera film footage indicators. Consists of really compact lever solenoid which actuates a pawl on a ratchet wheel to move a pointer progressively round a 24 volts d.c. and records 125 counts per revolution. Can be, used for mechanical counting display, etc., or the lever solenoid could readily be adapted for noid could readily be adapted for use in modelling. Housed in
diecast case with centre toggle diecast case with centre toggle on dial graduated $0-125$ supplied. 10/6 on dial graduated $0-125$ supplied. post


## THERMOSTAT BARGAIN

American Satchwell industrial quality type ST thermostat in neat strong plastic case size $1 \times 1 \frac{1}{8} \times 2 \frac{1}{8} \mathrm{in}$. deep. Heavy duty contacts rated at 15 amp . on $200 / 250$ volt A.C. ( $3 \frac{3}{4}$ kilowatts) adjustable by pointer in a recess at rear to break over a range of temperature around $70^{\circ} \mathrm{F}$. Recessed screw terminals at rear. Flush metal plate at (temperature sensitive) front, with two mounting lugs protruding fin. either side. Attractive red case.

Brand new $8 / 6$ post paid.
Brand New, Individually Tested, Fully Guaranteed LOW-VOLTAGE HALOGEN-QUENCHED, GEIGER-MUELLER TUBES 25'- post free
Working voltage 400-450. Highly sensitive. Effective length 11.8 cm . Background count 90 minute. Response 30,000 counts/minute. $80-$ volt plateau. Standard British 4-pin base, stainless iron electrode. Ideal for basic experimentation and instructional demonstration. Circuits of request with each tube.

## INVERTORS

28 Volt DC to $115 v$ I phase AC
Self-contained motor generator unit with complementary carbon pile voltage regulator, contactor and associated rectifier in separate compartment on same base. Continuously rated for $25 / 28$ volts D.C. input with 360 VA output at 115 volts single phase A.C. at 1,600 cycles with a power factor of 1.0. Fan cooled with end plate for blast or internal cooling as required. Type 200. Ref. 5UB/5083.

28 volt DC to $115 v 3$ phase $400 \mathrm{c} / \mathrm{s}$ AC. Type 102 A
Output 625 VA . Brand new. $\mathbf{1} 10$ carriage $10 /$.
200/220 Volt DC to $200 / 250 \mathrm{v}$ I phase $50 \mathrm{c} / \mathrm{s}$ AC
Output 260 watts. New, in soundproof cabinet. carriage paid. £9.10.0
24 volt DC to 26 v I phase $400 \mathrm{c} / \mathrm{s}$ AC
Output 6 VA . Size $2 \frac{1}{4} \mathrm{in}$. dia. $\times 4 \mathrm{in}$. long on $1 \frac{1}{2} \mathrm{in}$. high pedestal base. Instrument quality. As new $\mathbf{E l} 10.0$ carriage paid.

## ELECTRONIC ignition ANALYSER

A versatile, portable instrument, specially designed for use by the Services in the critical examination of aircraft engine ignition. systems during preventive maintenance routines.
It displays the entire performance of the ignition system on a cathode ray screen while the engine is running-simultaneously showing each plug's firing for side-by-side comparison together with any associated contact-breaker bounce, etc. Reveals excessive carbon formation, faulty condenser, leaking cables, wide gap, narrow gap, for either plug or contact-breaker points, worn cams, etc., and contains a 10 -step loading switch which absorbs energy from the ignition system and thus accentuates the test to show up deterioration in such items as coil primary or secondary winding, condenser, cables, etc., which will soon cause misfiring and eventual breakdown unless serviced or replaced.
Simply and quickly connected by crocodile clip leads to coil terminals and one "triggering" plug, the power supply can be switched from the front panel to suit either 250 volt A.C. mains, 6,12 or 24 volt battery, and can thus be used inside the car during road service if desired. In attractive, front and rear opening, black case, $9 \frac{1}{4}$ $13 \frac{3}{3} \times 17^{3} \mathrm{in}$. deep, weight only $33 \frac{1}{2} \mathrm{lb}$. Supplied complete with circuit, instructions, and good and faulty trace drawings. Guaranteed fully serviceable.

215 plus 10/- carriage.
I $\frac{1}{2}$ SECS. TIME DELAY SWITCH BRANDNEW Beautiful little, fast running, precision, ball bearing, double armature motor ( $1 \frac{1}{4} \mathrm{in}$. dia.) driving two spring returned cams through a precision ball bearing gearbox and magnetic solenoid clutch. The $1 \frac{1}{4}$ in. diameter steel cams have roller and tooth form followers on twin rocking arms that each operate switches with substantial changeover contacts and sprung microswitch action. A large-contact miniature relay is mounted under the rocker bridge. High quality modern unit made by Teddington Controls (TYPE FHM/A) controlling over-run of camera guns. Designed for $24 \mathrm{v} . \mathrm{D} . \mathrm{C}$. , but motor runs energetically at 6 v . when detached from gearbox. Totally sealed in robust die
cast box $4 \frac{1}{4} \times 3 \times 3$ in. with rubber cable seal protruding from flush mounted terminal plate. Time delay $1 \frac{1}{2}$ seconds. AM Ref. 5C/4435.

$$
35 /=\text { post paid. }
$$

## PRESSURE SENSING INDUCTANCE

Highly sensitive device consisting of a ferrite encapsulated $160 \mathrm{kc} / \mathrm{s}$ coil with a movable ferrite core attached to the free end of a singledisc aneroid capsule so that it transmits a change in frequency equivalent to the change in atmosphere pressure with increasing altitude. Coil Q, 43. Capacitance 870 pf. Housed in a $\frac{7}{6}$ in. square aluminium can on a lightweight $2 \frac{1}{2}$ in. diameter plug-in unit.

$$
\text { New } 25 / \text { - post paid. }
$$

## GROUND STATION TRANSMITTER

Type 75C, comprising RF Unit, RF Driver, RF Power Amplifier, Modulator, Modulator Power Unit, and Control Unit, all in 6 foot high 19 inch enclosed rack with full length rear access doors. This was the RAF ground station for operational communication with aircraft in the $100-150 \mathrm{Mc} / \mathrm{s}$ range and it is suggested that subst exceptional rig. Warehouse inspection invited.

Complete 235 carriage $£ 4$.

## TRANSMITTER / RECEIVER APN-I

A complete 14 -valve radar set covering $420-460 \mathrm{Mc} / \mathrm{s}$ ldeal for con version to radio control of models or 70 cm . work.
TRANSMITTER COMPRISES:
a push-pull feed-back oscillator tuneable either side of $445 \mathrm{Mc} / \mathrm{s}$. frequency modulated at $100 \mathrm{c} / \mathrm{s}$ by a particularly robust moving coil transducer. Two 955 high frequency acorn valves.
RECEIVER is tuneable to transmitter frequency. Two 9004 acorn valvesio AMPLIFIER
Self-contained RC coupled 12SH7, 12SH7 and 12SJ7. Amplifies the received signal which is passed to detector circuit giving a D.C. voltage proportional to the difference between the transmitted and received (reflected) signal to operate internal relays which pass appropriate correction signals to autopilot and supply external indicator ( 5 mA meter).

## MAIN CHASSIS

The main chassis carries the 3 sub-units and has a further three 12 SH 7 one 12SJ7, two 12 H 6 and one VR150 regulator.
BRAND NEW, a very useful buy indeed at only 22 plus $7 / 6$ carriage. less Dynamotor.

## PROOPS MECHANICAL OFFERS

## TACHOMETER CALIBRATOR

## MARK II

R.P.M. tester with direct and reduced ratio driving shafts and three ranges of speed indication by dual sensitivity galvanometer in Maxweil Bridge circuit. Heavy duty, 24 volt, 6 in . dia.s, ${ }^{3} \mathrm{~h}$ h.p. motor with coarse and fine speed control into $1: 1$ and $1: 4$ output drives giving 0 to 1,250 and 0 to 5,000 r.p.m. for testing direct and gearbox type tachometer generators. Interlocked forward and reverse switching. Ten-position speed selector for balancing bridge over each of three ranges 600 to $5,000,1,200$ to 10,000 and 2,400 to 20,000 r.p.m. Final balancing done at increased sensitivity by push button control. Quick mounting provision for two indicators and generators with two sets of quick fitting interconnecting leads, spare flexible drives, spare brushes, bulbs, etc., in rear compartment. Smart grey enamel bench unit with sloping pancl, overall size 19 in . high $\times 15 \mathrm{in}$. deep $\times 16 \mathrm{in}$. wide, plus 1lin. extension platform for generators.

Fully serviceable, carriage paid, $£ 20$

## SOLENOID OPERATED VALVES

Self-contained, precision engineered solenoid operated valves for use in gas or fluid flow control circuits. Completely sealed in cylinnections. For 12 or 24 volts D.C. operation. Brand new. $10 / 6$ HIGH-SPEED BEARING GREASE
post paid
10/6
Large tubes of high temperature, high-speed bearing grease by a famous manufacturer for use in aircraft landing wheel bearings, etc. Suitable for car wheel bearings, high-speed races, etc.

3 tubes for $5 /=$ post paid.

## PORTABLE STORAGE TANKS

 Brand new, high-duty, flexi-ble, aircraft fuel tanks. Made ble, aircraft fuel tanks. Made ally proofed, plastic material ally proofed, plastic material
impervious to oil, kerosene, water, etc. Capacity approximately 40 galls--can be folded into convenient carry$\begin{aligned} & \text { ing size when empty. Size: } \\ & 34 \mathrm{in} . ~\end{aligned} 282 \mathrm{in} . \times 7 \mathrm{in}$. tapering $34 i n, ~$
to
$4 \frac{1}{2}$ in.
Supplied fitted with submerged pump (described below),
SUBMERGED PUMP


## Oil Pressure Gauge

## and Transmitter

Ex-R.A.F. remote reading electrical oil pressure gauge. Transmitter is readily fitted to engine pressure connection. Circular scale, 2 -inch gauge, graduated $0-120 \mathrm{lb} / \mathrm{sq}$ in.

## 25/- post paid.

## LOW-PRESSURE WARNING SWITCH

Decreasing pressure on diaphragm closes switch contacts when pressure falls to a predetermined value-to operate warning lights, alarm bells, falls to automatically shut-off defective engine. Contact adjuster permits any setting between approximately 1 and $15 \mathrm{lb} / \mathrm{sq}$. in. Excellent protection device for marine or generating engine. 10/-
post paid.

## ELECTRIC ACTUATORS

Special offer of aircraft quality, precisioned engineered rotary actuators by leading British manufacturers. In new or first clas used condition. For 24 volt operation
Range:
TYPE 1


TYPE 2
Split field, series wound reversible motor fitted with electro-magnetic brake. Max. load 50/60 $\mathrm{lb} / \mathrm{ft}$. Output $0.02 \mathrm{~h} . \mathrm{p}$. at 13,000 r.p.m. Reduction gear ratio 2857 to 1 . Length 7 inches. Weight $2 \frac{1}{2} \mathrm{lb}$. Fitted with adjustable

Current 0.5 amps. Ram travel $90^{\circ}$. Maximum load $4.5 \mathrm{lb} / \mathrm{ft}$. Reduction gear ratio 4450 to 1 . Length overall $4 \frac{1}{2}$ inches. Weight 2 lb . $50 /-$ post paid.
limit switches 75/- post paid.
TYPE 3
Similar in appearance to above. Designed for operation of 3-position type valves in which actuator gives wide variety of angular settings Oetermined by position 0 of limit switches. Max. $0.017 \mathrm{~h} . \mathrm{p}$. at $17,000 \mathrm{r} . \mathrm{p} . \mathrm{m}$. Full range travel- 140 lb in 2 Output 0.017 h.p. at 17,000 r.p.m. F
seconds. Weight $3.25 \mathrm{lb} .75 /$-post paid.


Precision made, diecast-framed, centrifugal vane type pump. Intended for flange mounting through wall of tank-will pump fuel at $10 \mathrm{lb} / \mathrm{sq}$. in. at rate of 400 galls. per hour. Operated from self-contained, sealed motor rated 24 volts D.C. at 12.5 amp . Overall length 12 inches; flange diameter 8 inches. In excellent used condition and fully $\quad$ guaranteed $50 /=$ post


Two-pole, split series wound motor. Fitted with double-plate friction clutch. Speed of motor 11,000 r.p.m. epicyclic and worm epicyclic and worm gears to $60^{\circ}$ rotation of right-angled drive shaft in 3 seconds. Consumption 3 amps .

$75 /$ post paid.

## "KIGASS" PRIMING PUMPS

Very powerful hand-operated priming pumps constructed of nicke plated brass and fitted with inlet and outlet non-return valvesstandard BSP connections. Ideal for priming engines, hydraulic systems, large pumps, etc. Length: $10 \frac{1}{3}$ inches. $15 /=$ post paid.
Barrel diameter: $1 \frac{1}{4}$ inches. Similar, smaller type, with plain brass finish. $10 /=$ post paid.
Size: $8 \times$ in. dia.

## HYMATIC COMPRESSORS

High-quality, with machined cooling fins and detachable head. Compresses up to $450 \mathrm{lb} / \mathrm{sq}$. in. and delivers at up to 60 cu . in. per minute. Basis for home compressor/paint spraying outfits. Complet with extended splined driving shaft. $30 /=$
post paid.

## TOUGH RUBBER HOSE

Brand new, corrugated rubber hose with cord reinforced carcase and fabric covered exterior moulded to rubber to provide tough yet flexible tubing. Internal diamater ${ }^{\frac{3}{4}} \mathrm{in}$. inches.
In 3 foot lengths.
$2 / 6$

TYPE 4
Maximum load $35 \mathrm{lb} / \mathrm{ft}$ at 52 r.p.m. Clutch setting fevit. Reversible, spli field motor. Reduction gear ratio 275 to 1 . Length 87 inches. Width $4 \frac{3}{4}$ inches

75/.
75/- post paid.


## -

- 



## RCA AR88D RECEIVERS

One of the most renowned American Communications Receivers ever manufactured. Widely used by all the Armed Services to maintain World-wide Communications and Monitoring Posts under all conditions. Employs 14 valves, and has 6 switched overlapping wave ditions. Employs 14 valves, and has 6 switched overlapping wave
bands for complete coverage. Refinements include Mechanical bands for complete coverage, Refinements include Mechanical control. Automatic or Manual Noise Limiter, BFO with pitch control, RF and AF Gain Controls. Variable HF Tone Control, Variable Selectivity with Crystal Filter, Aerial Trimmer, Choice of Head phones or Speaker. Has internal mains power pack for nominal $15-230$ volts A.C. In Black Crackled Case size 194in. W. x 11 in . H.x 19tin. Ds Thoroughly reconditioned, immaculate in appearance, and in perfect working order. Covers $500 \mathrm{kc} / \mathrm{s}-32 \mathrm{Mc} / \mathrm{s}$, price (add carriage $30 /$ and $50 /$ - deposit on return
S.A.E. brings illustrated descriptive leaflet.

## TRAWLER BAND R 1155s

The latest version of this famous Communications Receiver to be released by the Air Ministry. Covers 5 wave ranges $18.5-7.5 \mathrm{Mc} / \mathrm{s}$., released by the Air Ministry. Covers 5 wave ranges $18.5-7.5$ Mc/s.,
$7.5-3.0 \mathrm{Mc} / \mathrm{s} ., 3.0-1.5 \mathrm{Mc} / \mathrm{s}, \quad 1.5 \mathrm{Mc} / \mathrm{s} ., 600 \mathrm{kc} / \mathrm{s} ., 500-200 \mathrm{kc} / \mathrm{s}$. As $7.5-3.0 \mathrm{Mc} / \mathrm{s},, 3.0-1.5 \mathrm{Mc} / \mathrm{s}$, , $1.5 \mathrm{Mc} / \mathrm{s}$, , $600 \mathrm{kc} / \mathrm{s}$,, $500-200 \mathrm{kc} / \mathrm{s}$. As
used by Coastal Command, Air-sea Rescue Launches, etc. All sets used by Coastal Command, Air-sea Rescue Launches, etc. All sets
thoroughly tested and in perfect working order before despatch, thoroughly tested and in perfect working order before despatch,
and on demonstration to callers. Have had slight use, but are in and on demonstration to callers. Have
A.C. MAINS POWER PACK OUTPUT STAGE in black metal A.G. MAINS POWER PACK OUTPUT STAGE in black meta case to match receiver, enabling it to be operated immediately, by just plugging in, without any modiscation. Fitted with 81 . CEIVER AND POWER PACK TOGETHER
Send S.A.E. for illustrated leaflet, or $1 / 3$ for 14 -page booklet which gives technical information, circuits, etc., and is supplied free with each receiver. Add carriage $10 / 6$ for Receiver, $5 /$ - for Power Unit.

## UNIVERSAL VOLT-OHM-

## MILLIAMMETER

Reads A.C. and D.C
Volts up to 1,000 in 5 ranges at 1,000 o.p.v., D.C. Current (3 ranges) to 500 mA . Resistance readings to 200 Kohms in $\begin{array}{cc}\text { ranges. } & \text { Basic } \\ \text { movement } & 300 \mu \mathrm{~A}\end{array}$ $\begin{array}{ll}\text { movement } & 300 \mu \mathrm{~A} \\ \text { sensitivity. } & \text { Easily }\end{array}$ sensitivity. open scale Dimensions 5fin. $x$ $3 \frac{1}{8}$ in. x 2 pin. Beautifully made, and fully guaranteed. Complete with leads, prods and internal bat:
tery. ONLY
$59 / 6$


HR.O. SENIOR COMMONICATIONS REGEIVERS. Complete with all 9 coils, giving coverage of $50 \mathrm{kc} / \mathrm{s}$ -
$30 \mathrm{Mc} / \mathrm{s}$. Checked and in perfect working orier. Rack mounting type 18 gns., standard Table Model 21 gns. (Carr., etc. 22/- either type.)
HRO MAINS POWER UNITS, A.C. Input $115 / 230$ volte, Output D.O. (fully smoothed) 230 volts 75 mA ., and
6 .. orts 3.5 amps. Complete in black crackled case
ONy ONLY 59/6.
BC 342 RECEIVERS. A few only of these famous American scts covering $1.15 \cdot 18.0 \mathrm{Mc} / \mathrm{s}$, , in six bands. Internal condition and perfect working order. ONLY $£ 25$ (carriage 15/0).
12-WAY SCREENED CABLE, In 10ft. lengths, fitted with plugs, origtnally made for No. 19 Wireless set鲑

JNISELECTOR. 25 way 8 bank 15 onm coll for 24 volt operation, unused ez-brand new equipment, oniy $62 / 6$. SPRAGUE CONDENSERS. Metal cased wire ends. New, $1 \mathrm{mfd} .1,000 \mathrm{v}$. and .1 mid. 500 v . $7 / 8$ per dozen. Special quotes for quantities.

## HETERODYNE FREQUENCY METERS TYPE LMI4



Frequency range $125-20,000 \mathrm{kc} / \mathrm{s}$. in 2 bands. This is the United States Navy Model of the well-known BC. 221 Frequency Meter, but has many alditional features Which increase its usefiness. Voltage stabilisation cir addition it is titted with an Internal Modulation switch o allow use as a Signal Generator. Size oniy 8 in. $\times$ Sin . $\times 8$ 8공in. Full information on request.

## BC 221 FREQUENCY METERS

Similar specification to LM 14 Frequency Meter below, but does not have internal modulation or voltage stabilising circuits. Complete with original calibration book, crystal, valves, and instruction book. Used, but in very good condition. ONLY £16. Illustrated descriptive leaflet available on request.

## 20,000 OHMS PER VOLT TESTMETER

ust purchased from the Alr Ministry, these magnificent American PRECISION Testmeters provide 31 ranges for reading Voltage Current and Resistance, 6 Decibel
ranges, 7 Output ranges, and faclitice for teating Electroranges, 7 Output ranges, and facilities for testing Electro ytic and Paper Condensers. Single wwlich control, mounted on Black and Silver Panel size Byin. $x 7$ in which is fitted into turdy wool carrying case with removable hinged lid and compartment for leads, etc. the overall size being $10 \mathrm{in} \times 9 \mathrm{in}, \times 6 \mathrm{in}$. Has 7 D.C Voltage ranges up to 6,000 volts at 20,000 ohma per polt simillar ranges at 1,000 ohms per volt, 7 similar A.C. hought micrormps to 12 Amps., 3 Resiatance ranges ap to 60 Megohms, 6 Decibel ranges from -12 to +70 DB, 7 Output ranges up to 6,000 rolts. Supplied with est leads, internal batteries, and operating instrictions. Case finished medJum Oak, and bitted with leather handle. In excellent condition, thoroughly checked before des-
patch. 40 only avaitable- first come first served. ONLY $89 / 19 / 6$ (post etc. $5 / 6$ ).
Further details on receipt of S.A.E.

## DOUBLE BEAM

## OSCILLOSCOPE TUBES

Type CV 1690 equivalent to Cossor O9D as used in oscilloscopes by Cossor ( 339 series). Hartley and £2/19/6 (carriage 5/6). Brand new in makers' prate

## METERS

$\left.\begin{array}{rrrr}\text { F.S.D. } & \text { SIZE AND TYPE } & \text { PRICE } \\ 25 \text { microamps } & \text { D.C. } & \text { 2in. Projr } & \text { circular }\end{array}\right) 599 / 6$

HIGH FREQUENCY A.C. VOLTMETER


AMPLIFIER N84. As previously advertised. 4 valves ack mounting, with internal A.C. mains pack for nominal
$10 / 230$ volts. Output to 600 ohms. Ine, provision for 00 ohms. or Hixh Impedance foput. A first class fob NEW IN MAKERS' PACKING. ONLY $89 / 6$ (carr. 10/6)

INTERCOM. TELEPHONE SET. Two pairs of Brand New Headphomes connected to Breast Microphones, 4f volt battery, 10 yards twin flex, and fult instructions for connecting to make super intercom. ONLY 2 //6. (Post 3/6). Extra flex 3u. per yard.

10,000 OHMS PER VOLT TESTMETER. This latest Caby model is a handy pocket sized tester of $\times 3$ caby model is a handy pocket sized tester of $\times 33$
$\times 2 t i n$. Reads low D.O. voltages at 10,000 ohms per volt, up to $10,000 \mathrm{v}$. A.C. and D.C. at 4,000 o.p.r. Resistance to 20 megs. D.C. current to 250 nilliamps and also Decibels. Complete with Test Leads, Batterice and Instruction Book. ONLY £6/10/-

12 VOLTS AMERICAN DYNAMOTOR. Delivers 220 volt at 100 mills. Size $5 t \times 3$ in. diameter. Ideal for runnlag Radio and Electric shaver etc., from car battery. ONLY 32/6.

MARCONI SIGNAL GENERATOR TF 144G/7. Coverage $5 \mathrm{kc} / \mathrm{s}=2.5 \mathrm{Me} / \mathrm{s}$. and $8 \mathrm{Mc} / \mathrm{s} .-70 \mathrm{Mc} / \mathrm{s}$. Complete, and In A8 NEW CONDITION. ONLY £95

## "P.W." 6 TRANSISTOR PERSONAL RECEIVER



Designed by the technical staf
of Practical of Practical
Wirelass, easy to build, usin Hand 1st Grad Matched Tran Full Medium coverage to inter nal speaker. All rately (new components only) enabling you to buy as required, and fuil detailed price list will be sent on request. BATTERY AND CABINET. £8/19/6.

Cash with order please, and print name and address clearly PLEASE ADD POSTAGE OR CARRIAGE COSTS ON ALL ITEMS

## HARRIS ELECTRONICS (LONDON) LTD.

Radio Corner, 138 Gray's Inn Road, London, W.C.1. Phone: TERMINUS 7937

Open until 1 p.m. Saturdays.

We are 2 mins. from High Holborn (Chancery Lane Station) and 5 mins. by bus from King's Cross
 Ferrite rod internal zerial, high flux $7 \times 4 \mathrm{fin}$ Loudspeaker. Cabinet wilh first quality walnut veneer finisb and gold embellishmente. Slue: $18 \times 8 \frac{1}{2} \times 5 \mathrm{~m}$.

Circult diagram and full data supplied. Every component available separately.

SAVE POUNDS! ORDER BY POST IF YOU CANNOT CALL

## LASKY'S FOR MINIATURE POCKET TRANSISTOR RADIOS!

Add to the pleasure of Summer outings, holidays, camping, etc. Wide choice of well-known makes at keenest prices. Call and select or order by post. A few examples.

FIDELITY 6 LEECO 6 SHIRA 6 PERDIO 6 CROWN 6 EMERSON 6 DANSETTE 6 PHILIPS 7 BRAUN 7

## $9 \frac{1}{2}$ Gns.

 91 ${ }^{2}$ Gns. 91 Gns. $10 \frac{1}{\mathrm{G}} \mathrm{Gns}$. $11 \frac{1}{2}$ Gns. 11.19.6. 14 Gns. 15 Gns. 15 Gns.15主Gns.

Regd. postage and insurance $5 /$-.

4-SPD AUTO-CHANGERS
New and Unused in Makers' Cartons.
B.S.R. type UA8 ...... \&6 196 B.S.R. type UA8 B.S.R. UA12, stereo B.S.R. Type UA14 B.S.R. Type UAl4 COLLARO Studio c6̈o \& 8196 stereo, with monaural \& $19 \frac{19}{6}$ Post on all above $5 /$ -

GARRARD

## Model 120

Model 121
Model 209
Mdl. 210, Stereo $\begin{array}{rr}£ 8 & 8 \\ £ 9 & 9 \\ £ 9 & 19 \\ £ 11 & 0\end{array}$ Mdl. 210 with
monaur RC. 88
RC. 88 STEREO
RC. 88 .
TAPE RECORDER BARGAINS

## SINGLE PLAYERS

Auto start and stop. Complete with pick-up and crystal cartridge. GARRARD 4SP GARRARD TA Mk. II, wired for STEREO, plug-in head \&s 90 E.M.I. 4 -spd., wired for STEREO and fitted mono car-
tridge ................. 196 As above, STEREO...... ${ }^{〔 6}$
Post on all above $5 /$. COLLARO JUNIOR 4 -specd motor and separate pick-up
B.S.R. TU9, non-aut B.S.R. TU9, non-auto Turntable and separate pick-up

PICK-UP CARTRIDGES
ACOS HGP. 59 or HGP. 37 turn over Acostal cartridge with L.P. and crystal cartridge with L.
standard styli. List $39 / 7$.
Lasky's Price 18/- post free.
$\overline{\mathrm{ACOS}} 73$-1A STEREO. List $52 / 6$. Lasky's Price 29/6 post free.
MICROPHONE BARGAINS ACOS CRYSTAL STICK MIKE Type M.C.39/1, complete wit

Crystal Hand or Table Mike. 15/Post free.
MINIATURE moving coil dynamic microphone, incorporating switch and pocket clip. As used for the " Fi-Cord" 35/- Post $1 / 6$.

## HIGH FIDELITY TAPE <br> RECORDER HEADS

Leading make, new and unused. Upper
or tower track. RECORD/PLAYBACK
high tompedance. high imperdance. Double wound and wid
reproduce ap to 12,000 c.p.e. at $7 t$ Azimuth adjustments. C.p.c. atput ${ }^{5}$ i.p.is. voits at 1 Kc . at $7 \neq$ l.p.s. ERASE, low voits at kc . at

LASKY'S PRICE, per pair | Post free. |
| :--- |

Please specify upper or lower track.
SPECIAL OFFER. Set of 4 Heads (upper \& lower) 49/6


ELIZABETHAN "BANDBOX " for A.C. mains 200/250 v ., fitted fully self-contained Amplifier and $7 \times 4 \mathrm{in}$. Speaker. Clock type face indicator, monitoring and 1.s. sockets. 2 -speed, 3 and $1 \frac{1}{8}$ i.p.s., fast forward and fast rewind. Record level indicator. Facilities for recording from two inputs. Push-button controls. Plays one hour on one reel of tape. Case, $10 \frac{1}{2} \times 9 \times 6 \mathrm{in}$., with detachable hinged lid. LIST 29 Gns.
LASKY'S PRICE, with high quality crystal Mike and one reel of Tape, \&15.19.6 Carr. \& Ins. 15/-.

ANOTHER RECORDER BARGAIN! Well-known make using Collaro Studio 3 -speed deck, $1 \frac{2}{d}, 3 \neq \frac{3}{2}, 7 \frac{1}{2}$ i.p.s. Twin track with pause control, rev. counter, latest type electronic recording indicator. Superimposing rev. counter, latest type electronic record $\quad 4$ Speaker. Takes 7 in . spools. switch, volume and tone controls, $\times$ watts output. Contemporary design carrying Case, $9 \frac{1}{2} \times 16 \times 16 \mathrm{in}$. COMPLETE with Mike, Tape and Spool, 29 GNS.

## "SHERWIN SIX" <br> TRANSISTOR <br> POCKET SUPERHET

The very latest printed circuit, using six matched top grade S. T C. using sistors and germanium diode. Pushpull output feeding 3in. P.M. speaker. Full medium and long speaker. Internal ferrite aerial and provision for car aerial. Housed in provision for car aerial. Housed in size $6 \frac{1}{4} \times 4 \frac{x}{} 1 \frac{3}{3}$ in. Full point-topoint instructions supplied.

GAN BE BUILT FOR $\$ 8 / 19 / 6$ All components available separately. Circuit diagram and instruc tions, $2 / 6$ (refunded if you order).

> PLASTIC TAPE SPOOLS
> $\begin{array}{ccccc}3 \mathrm{in} . & 5 \mathrm{in} & 5 / \mathrm{in.} & 7 \mathrm{in} . & 8 / \mathrm{in} . \\ 1 / 9 & 2 / 6 & 2 / 6 & 2 / 6 & 5 / 6\end{array}$


The "CLARION" TRANSISTOR BATTERY TAPE RECORDER
Capstan drive, push-button controls. Constant speed $3 \frac{3}{8}$ i.p.s., uses 3in. spools. High impact plastic case with transparent upper. Size: $9 \frac{1}{2} \times 5 \times 3 \frac{1}{2}$. List 25 Gns. Lasky's Price, with Mike and Tape
Carr. 7/6.
18 GNS.

## TRANSISTOR RECORD PLAYER CAN BE BUILT FOR £9.19.6

6 volt operation. For all L.P. and standard records. Complete parcel comprises:AMPLIFIER. 300 milliwatts output, using two OC71 and two OC72 transistors. Fully assembled, 79/6. Knobs 3/6 extra. LOUDSPEAKER. 30 ohms, $7 \times 4$ in. elliptical Speaker matched to amplifier. $25 /$-.
3-SPEED TURNTABLE 3-SPEED TURNTABLE with rubber mat and speed adjustment, complete with i.o. crystal cartridge and two sapphire styli. 79/6
CARRYING CASE as illustrated, handsome two-tone finish, size 17 in . deep, 14in. wide, 5 in. high. 49/6.
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 $38.88 \quad 43 \quad 70 \quad 60 \quad 61.11 \quad 62 \quad 66 \quad 70 \quad 72: 22 \quad 8083.33 \quad 85 \quad 879093$ $\begin{array}{lllllllllllllllllll}94.44 & 97 & 6700 & 03.75 & 05.55 & 08 & 10 & 14 & 16.66 & 17 & 19 & 20 & 27.7\end{array}$ 30384045505354.7555566061 .116265708083 .3 $869094.4495 \quad 680005.551016 .6620222527 .77 \quad 3032$ $\begin{array}{lllllllllllllllllll}35 & 37 & 38.88 & 40 & 50 & 60 & 61 & 63 & 65 & 70 & 72.27 & 73 & 75 & 80 & 83.33 & 90\end{array}$ $\begin{array}{llllllllllllllllll}92.5 & 94.44 & 97.5 & 6900 & 02.5 & 05.55 & 10 & 16.66 & 20 & 27.77 & 30\end{array}$ 3538.8840434750566061 .11637072 .2280859092 .5 $94.495 \quad 7000 \quad 02051020223038.8840505560627073$ $75809095 \quad 710010203040485054606670728083$ $90 \quad 7200102025404550536070809095 \quad 73000405$ $10202530405060707375808590 \quad 74000510204050$ $\begin{array}{lllllllllllll}57 & 60 & 708090 & 98 & 7500 & 10 & 15 & 20 & 30 & 40 & 50 & 60 & 80 \\ 82 & 90 & 99\end{array}$ 7600020510192026303840505459677076778090 $770010203040505960707779809099 \quad 780010152030$ $404750607090 \quad 7900051020304050608090 \quad 8000$
 $\begin{array}{llllllllllllll}23.64 & 25 & 27.14 & 30 & 35.38 & 36.25 & 40 & 44 & 48.61 & 49.23 & 50 & 50.59\end{array}$ $\begin{array}{lllllllllllllllll}52.86 & 55.5 & 56.36 & 60 & 61.18 & 63.08 & 65.71 & 68 & 70 & 72.73 & 76 & 77\end{array}$ 78.5778 .68081 .258489 .0989 .259090 .7791 .43928100 03.5304 .2904 .304 .620507 .1410151617 .1418 .52026 .25 3032.3135 .335 .737 .537 .74042 .542 .8645 .8846 .1546 .16 $46.56 \quad 48.75 \quad 54.55 \quad 55.70 \quad 55.71 \quad 56 \quad 60 \quad 61.54 \quad 64.54 \quad 66 \quad 67.06$ 687071.2573 .85757677 .658081 .4387 .278892 .7593 .75 $\begin{array}{llllllllllllllllllll}94.29 & 95.45 & 98.82 & 8200 & 01.54 & 03 & 05 & 07.14 & 17 & 17.5 & 20 & 22.2\end{array}$ 242527.5283032 .6635 .883638 .46404445 .745 .7148 $\begin{array}{lllllllllllllllll}49.46 & 50 & 52.73 & 53.85 & 54 & 55 & 56 & 56.92 & 57 & 58.57 & 60 & 69.09 & 70\end{array}$ $\begin{array}{llllllllllllllll}71.43 & 72.5 & 77 & 78.57 & 80 & 83 & 84.29 & 84.62 & 85.45 & 89 & 90 & 92.14\end{array}$ $\begin{array}{llllllllll}94.12 & 95 & 96 & 97.14 & 99.83 & 8300 & 01.82 & 04.7 & 07.69 & 08 \\ 10 & 12.31\end{array}$ 15.3816 .717 .520222628303233 .334 .5535 .713638 .6 38.64404747 .0649 .46505253 .8555566061 .5462 .564 $67.69 \quad 67.76 \quad 68 \quad 7074.29 \quad 7577.88081 .54848587 .1490$ $\begin{array}{llllllllllllll}91.82 & 92 & 8400 & 04 & 05 & 09.23 & 10 & 12.86 & 15 & 16.15 & 20 & 22.5 & 23\end{array}$ $23.08 \quad 28 \quad 32.73 \quad 36.924041 .2545 \quad 5051.435252 .5 \quad 55 \quad 64.62$ $\begin{array}{llllllllllllll}65.45 & 65.46 & 69.23 & 70.46 & 71 & 76 & 77.14 & 78.46 & 80 & 81.82 & 84.62\end{array}$ $\begin{array}{lllllllllllll}88 & 90 & 92.31 & 97 & 98.18 & 8500 & 10 & 12 & 15.71 & 20 & 25 & 28 & 30.91\end{array}$ 404141.4344 .447 .2754555663 .64656667 .147070 .8 $768091.6696 .36 \quad 860012.732029 .09304045 .454761 .82$ $\begin{array}{llllllllllll}66.66 & 66.67 & 70 & 78.18 & 94.55 & 8700 & 04.29 & 10.91 & 44 & 45 & 56.7\end{array}$ $\begin{array}{llllllllllll}83.1 & 87.5 & 90 & 8800 & 85 & 8900 & 10 & 25 & 55 & 60 & 85 & 9012.5\end{array} 62$ $\begin{array}{lllllllllll}65 & 70 & 9125 & 50 & 9222 & 38.89 & 60 & 73 & 77.7 & 80 & 9315\end{array} 9400$ $\begin{array}{llllllllllll}28 & 44.4 & 55.55 & 88.9 & 9500 & 05.5 & 10 & 22.2 & 33.3 & 33.33 & 55.50\end{array}$ $\begin{array}{lllllllllll}66.66 & 88.8 & 88.9 & 9600 & 17.8 & 20 & 22.2 & 27 & 33.33 & 50 & 60 \\ 63 & 66.66\end{array}$

 $\begin{array}{lllllllllllllllll}55.5 & 55.55 & 77.7 & 83.3 & 88.88 & 88.9 & 9900 & 11.1 & 22.2 & 33.3 & 34 & 35\end{array}$ $\begin{array}{llllllllllll}40 & 44.44 & 55.55 & 77.8 & 10166 & 189 & 223 & 245 & 300 & 391.6 & 445\end{array}$ $477.8478488 \quad 501511534545556567622767800823$ $856878 \quad 11463501526587751788814851876$.
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120 and 121.

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Capable men with good qualifications are required for a variety of interesting posts in the field of Production Engineering, Development, and Research at salaries ranging from $£ 1,000$ per annum upwards.
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Suitably qualified Engineers are invited to write to the
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Applicants with a knowledge of this type of testing, and wishing to obtain well-paid positions, should apply stating qualifications and experience to the Personnel Manager, Box No. 5033, c/o "Wireless World."



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## UNITED KINGDOM ATOMIC ENERGY AUTHORITY PRODUCTION GROUP INSTRUMENT MECHANICS

Windscale and Calder Works, and Chapelcross Works require experienced men with knowledge of electronic equipment and/or industrial instrumentation for fault diagnosis, repair and calibration of a wide range of instruments used in nuclear reactors, radiation laboratories and chemical plant. This interesting work involves the maintenance of instruments using pulse techniques, wide band low noise amplifiers, pulse amplitude analysers, counting circuits, television and industrial instruments used for the measurement of pressure, temperature and flow.
Men with Services, Industrial or Commercial background of radar, radio, television, industrial or aircraft instruments are invited to write for further information. Training Courses in Specialised Techniques are provided for successful applicants having suitable Instrumentation background.

Married men living beyond daily travelling distance will be eligible for housing. A lodging allowance is payable whilst waiting for housing. Working conditions and promotion prospects are good.

Applications to:
Works Labour Manager, Windscale and Calder Works, Sellafield, Seascale, Cumberland
or
Works Labour Manager, Chapelcross Works, Annan, Dumfriesshire, Scotland.

# UNITED KINGDOM ATOMIC ENERGY AUTHORITY 

## dounpeay Expermental <br>  reactor establishuent <br> INSTRUMENT MECHANICS <br> (Physical and Electronic) <br> and INSTRUMENT ELECTRICIANS

There are vacancies in the Instrument Department for men with experience in the maintenance of instruments for the measurement of pressure, flow and temperature, electronic instruments, radar and television and for electricians with experience in the maintenance of temperature recorders and electromagnetic relays.

Applications are invited from men with experience of instruments in industry or with appropriate experience in H.M. Forces.
The rate of pay is $£ 1370 \mathrm{~d}$. for a 42 hour five day week and there is a superannuation scheme. Housing will be made available to married men and there is accommodation for single men and married men awaiting housing.

Facilities are available for further education and promotion prospects are good.

Application forms and further information can be obtained from:
Recruitment Officer, Dounreay E.R.E.,
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## BOURNEMOUTH MUNCIPAL COLLEGE OF TECHNOLOGY AND COMMERCE

Full-time and Part-time Courses for the INBTITUTION OF ELEGTRICAL ENGINEER8
PART III EXAMINATIONS
in
Advanced Electrical Engineering: and special subjects of both Heavy Current and Light Current Groups, are due to commence in September, 1961. Further details from: The Principal, Municipal College of Technology and Commerce, Lansdowne, Bournemouth. Early application is advisable.

## GRANADA TV

have vacancies at their London Studios for experienced TELEVISION BROADCASTING ENGINEERS.
Applicants without previous experience but with experience but with a good knowledge of electronics will also be considered
Five-day week, generous Pension and Life Assurance arrangements.
Applicants should write giving details of qualifications, exper ience and present employment to Personnel Officer, Granada to Personnel Network Limited, 36, Golden Square, London, W.I.

OPPORTUNITIES FOR SCIENCE AND WATHEMATIG8 GRADUATES TO TAKE UP TEACHINE
Graduates in Science and Mathematics who are having second thoughts about the careers they chose after leaving the university, or married women with degrees in these subjects who would like to teach full/ or part-time, should write to the Chief Education Officer (Dept. W.2), County Offices. Chelmsford, Essex, for particulars of a brief training scheme to make it easier for them to take up teaching.

If accepted for service in Essex schools, they can be given a short course of training and teaching practice before they take up and teaching practice before they take up
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INDEPENDENT TELEVISION BIRMINGHAM STUDIOS
Technical Assistants and Assistant Engineers are required at the above studios to fill the following posts:-

1. For duties concerned with the maintenance of TV studio equipment and the development of new devices.
2. For operational work on TV signal processing equipment. including film projectors and cameras but not transmitters, coupled with maintenance of the equipment. This post involves shift working and training will be given where necessary
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THE CHIEF ENGINEER,
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THE PEMBRIDGE COLLEGE OF ELECTRONICS offers training in RADIO TELEVISION AND ELECTRONICS

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Full-time One Year Course in Radio and Television. College course in basic principles for prospective servicing engineers.
Next course commences 5th September 1961
This course is recognised by the Radio Trades Examination Board (R.T.E.B.) for the new Servicing Certificate examinations.

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For details, write to:
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THE PEMBRIDGE COLLEGE OF ELECTRONICS

34a Hereford-Road, London, W. 2

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CYPHER OPERATORS MALE AND TELEPRINTER OPERATORS FEMALE Write, giving details of education, qualifications and experience, to:Personnel Officer, G.C.H.Q. (RCO/3)
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## SOLARTRON TEST ENGINEER

is required by our Test Department for the fault finding and testing of electronic instruments to government specifications.
Applicants should have had previous experience in this work, or dealt with radar in the Services or have serviced televisions and radios.
Ref. No. $577 / \mathrm{WW}$.
Please apply:-
John Delfgou,
Assistant Personnel Officer,
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Young men with good technical training and some knowledge of Telecommunications, Radio and Radar, are required for training as Telecommunication Installation Engineers. Candidates must be prepared to travel anywhere in U.K. or abroad. The situations offered are progressive and pensionable and offer good scope to suitable men desirous of making a career in Telecommunications.
Commencing salary from $£ 13$ to $£ 16$ per week plus a living allowance not less than $£ 5$ per week when working away from home.

Please write details of personal circumstances and career 10:-
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North Woolwich, London, E.16.

## ASSISTANT ENGINEERS (RADIO) URGENTLY REQUIRED BY DIRECTORATE OF CIVII. AVIATION <br> EAST AFRICA HIGH COMMISSION <br> Appointment on contract for 1 tour of 36 months in first instance. Salary according to age and experience in scale (including Inducement Pay) rising to $£ 1,671$ a year. Gratuity $13 \frac{1}{2} \%$ of total salary drawn. Outfit Allowance $£ 30$. An Education Allowance ranging from $£ 50$ to $£ 200$ is likely to be introduced shortly. Free passages. Liberal leave on full salary. second year and preferably one third year C. \& G. Certificate in Telecommunications Principles or Radio and have had least 3 years' experience in arection and meintenance of around station transmitters, radio and radar navigational aids and aerial station t. systems. <br> Apply to CROWN AGENTS, 4 Millbank, London, S.W.1, for application form and further particulars, stating age, name, for application form and further particulars, stating age, name, brief details reference $M 2 A / 51200 / W F$

## ELECTRONIC DEVELOPMENT ENGINEER

required
to lead a small team engaged in the development of commercial communication equipment. Applicants should have a degree or equivalent qualification and some years experience as Project Engineers. Preferred age range: 30/35 years.
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Write giving details of education, qualifications and past experience to:-

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Applications are invited from young men for maintenance of large-scale computer installations. The work involves testing, fault diagnosis and maintenance of a large volume of fully transistorised electronic circuits.
Applicants, shouid have a degree, H.N.C. or similar qualification together with some practical experience of electronic work. Men with appreciable experience but without qualifications will be considered.
Comprehensive training will be given and excellent prospects exist in the Company which is expanding rapidly.
Details of experience, qualifications, age and present salary should be sent to: Personnel Manager, LEO Computers Limited, Hartree House, 151A-159A Queensway, London, W.2.

## TELEPHONE CARRIER ENGINEER

International Aeradio Limited invite applications from engineers with 7-8 years practical experience in the design, planning and field installation of radio carrier telephone equipment of radio carrier telephone equipment of
which at least 3 years must have been in a supervisory capacity. Experience in a supervisory capacity. Experience of small manual and automatic telephone exchanges would be an added advantage. The successful candidate will be required to serve initially in Indonesia for a period of 3 years, during which a tax-free salary will be paid together with living, separation and kit allowances; free accommodation will
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40 Park Street, London, W.1.

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SENIOR \& JUNIOR design and development engineers
required for work on Electronic Meosuring Equipment, connected with Production Testing of Communications Equipment, Radio, Television Equipment and Components on a semi-automatic basis, probably leading to automation.

Applicants must have technical abilities; a keen interest in new techniques and methods, associated with Production Testing on a large scale, and must be suitably qualified Electronics Engineers or have had considerably practical experience in Electronic Measurements. A mechanical background or interest such as model engineering would be an advantage.

The appointments carry generous and progressive salaries commensurate with abilities and experience.

- Applications which will be treated in confidence should be addressed for the ottention of The Personnel Manager, The Plessey Company Limited, Vicarage Lane, IIford, Essex.


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 EleCTRICAL ENGINEERS for work on
## CONTROL SYSTEMS

We invite applications from engineers to join our project teams working on the design of flight control systems, engine controls, and servomechanisms, and pulse techniques.
There are vacancies for graduate engineers, for those with ordinary and higher national certificates and for those without qualifications but with some experience in these fields. The work will be carried out at the Company's Main R and D establishment at Hatfield, where all facilities exist for the development of high efficiency equipment.
Replies in confidence, should be addressed to:-
The Personnel Manager, (Ref. 187).
THE DE HAVILLAND AIRCRAFT COMPANY LIMITED
Hatfield, Herts.
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## RADIO POLICE

An INSPECTOR OF POLICE (SIGNALS) required by GOVERNMENT OF NYASALAND.

Choice of contract terms for 1 tour 2-3 years with $10 \%$ gratuity or permanent and pensionable terms. Commencing salary according to age and experience in scale rising to $£ 1,285$. Cost of Living Allowance of $5 \%$ of salary also payable. Initial outfit grant and annual uniform allowance. Free passages. Liberal leave on full salary.
Candidates of good education and physique, normal vision without glasses, must have sound knowledge of H.F. and without glasses, must have sound knowledge of H.F. and V.H.F. fixed and mobile simplex and duplex radio telephone
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Apply to CROWN AGENTS, 4 Millbank, London S.W.1., for application form and further particulars, stating age, name, Brief details of qualifications and experience and quoting reference M2A/50901/WF.

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Required for the London Office of the CROWN AGENTS FOR OVERSEA GOVERNMENTS AND ADMINISTRATIONS for avpointment normally to pensionable establishment on probation for appointment normally to pensionable establishment on probation for
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Candidates, preferably 25-35 years of age. should have a degree in Electrical Engineering or be Graduate or Corporate Members of the Institution of Electrical Engineers. They should have received a thorough training in Automatic Telephone Exchange and associated equipment with a telecommunications manufacturer or Authority and have had subsequent operating or design experience.
Duties include purchase of telephone equipment, preparation of specifications for tenders, adjudication of tenders and technical correspondence with Administrations.
Apply to CROWN AGENTS, 4, Millbank, London, S.W.1, for further particulars, stating age, name, brief details of qualifications and further particulars, stating age, name, briet details
experience and quoting reference M2A/51284/WF.

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## INSTALLATION \& MAINTENANCE ENGINEERS

EMI Electronics Ltd. is expanding its production programme and accordingly has vacancies for a number of installation and maintenance engineers. The work involves the latest electronic techniques covering a wide range of equipment which can be grouped as follows:
(1) Instrument tape decks, electronic balancers, stroboscopes
(2) Machine tool control, analogue computers, process control systems
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(4) Colour TV, film scanners, studio cameras.

Applications from experienced television service engineers and ex-service radar personnel are particularly invited. These programme positions carry staff status and are pensionable.

Please write, giving full details and quoting Ref. EL/30/7, to:

> Personnel Manager, EMI ELECTRONICS LTD. HAYES, MIDDLESEX.

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 Eastern DivisionApplications are invited for the following permanent superannuable post in the ing permanent superannuable post in the ated at Hornsey (NORTH LONDON).

ELECTRONIC MAINTENANCE ASSISTANT
Salary range $£ 830-£ 1,065$ per annum
The work is interesting and varied and involves servicing instruments employing involves servicing instruments employing
wide-band amplifiers, counting circuits and the latest techniques in pulse height the latest
Applicants should have several years ${ }^{*}$ experience in the maintenance of industrial experience in the maintenance of industrial
or domestic electronic equipment and or domestic electronic equipment and
preferably be familiar with pulse techpreferab
niques.
niques.
Preference will be given to those possessing a Higher National Certificate or equivalent qualifications.
Applications, quoting reference S.V. No. 1449, stating age, qualifications, experience and present position, should be pentence the present position, should be sent to the Controller, Central Electricity Generating Board, Easterm Division, West Farm Place, Chalk Lane, Cockfosters, Barnet, ferts., to
arrive not later than the 3 d Junc, 1961 .

INTERNATIONAL

## AERADIO

 LIMITEDrequire
RADIO INSTRUCTORS
to be employed at their new Training School now in course of erection. Two Schooncies exist, as follows:-

1. TECHNICAL INBTRUCTOR. Should possess a recognised professional qualifica tion in telecommunications, with previous instructional experience. Familiarity with a range of telecommunications equipment is essential.
2. PRAGTICAL INSTRUGTOR. Should possess a background in telecommunications with previous instructional extions with previous instructional experience. Must show a high degree of kill and knowledge of practical tech niques as applied to telecommunication engineering, including assembly, installation and maintenance.
Salary will be commensurate with qualifications and experience and excellent working conditions apply The posts are permanent and pensionable. Applications to:The Personnel Officer,
INTERNATIONAL AERADIO LIMITED, Hayes Road, Southall, Middlesex.

FACANCIES FOR RESEARCE AND DEVELOPMENT CRAFTSMEN IN GOVERNMENT SERVICE Experience in one or more of the following:ELECTRICAL (1) Matntenance of radio communication recelvers
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Inatrument makers and genera machinitsts with bench futting and machine shop experlence for contype electronlc equipment
basic pay E9 188. 2d. plus merit pay in the range of 101 . to $100 /$ - per week. Merit pay will be assessed at interview, based on ability and the qecesary basic quali
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ionsble pasta Five-dual permanent and pen conditions; single accommodation available.
Apply in writing to: Personnel Offcer, G.C.B.Q.
(RDC/3), 63, Clarence Street, Cheltenham, Glos.

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Southampton County Borough
Education Committee
Southampton Technical College
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The College offers a three years' full-time course in Communication Engineering and Electronics which provides a comprehensive training in these subjects and leads to the College Diploma in Communications and Electronics. This Diploma carries complete exemption from the Graduateship Examination of the British Institution of Radio Engineers.

Further details of the course, fees, etc., obtainable from the Registrar, Southampton Technical College, St. Mary Street, Southampton.

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Phone: Cherrywood 3055

## RADIO POLICE NORTHERN RHODESIA

Radio Technicians required for appointment as INSPECTOR/ SENIOR INSPECTOR OF POLICE, NORTHERN RHODESIA, on agreement for one tour of 3 years in first instance with prospects of permanent and pensionable employment. Salary according to age and experience in scale rising to $£ 1,380$ a year. Plain clothes allowance £24 a year. Married accommodation with heavy furniture available immediately at low rental. Free passages. Liberal leave on full salary.

Candidates, 23 to 35 years of age, of good physique, should possess maths and physics at G.C.E. "O" level standard. They should have sound knowledge of installation and maintenance of modern low and medium power V.H.F. static and mobile equipment, H.F. transmitters and receivers, including S.S.B., and petrol generator and diesel electric sets. Knowledge of installation and maintenance of teleprinters would be an advantage. Apply to CROWN AGENTS, 4 Millbank, London, S.W.1, for application form and further particulars, stating age, name, brief details of qualifications and experience and quoting reference M2A/51291/WF.

## ELioT

ELECTRICAL \& ELECTRONIC ENGINEERS

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are required in the Transport Aircraft Controls Division for interesting long-term projects connected with

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The work involves the application and development of the latest techniques in the design of transistor circuits, servos and integrated systems for present and future aircraft.

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Applications, stating qualifications, age and experience, should be addressed to:

The Personnel Manager (Ref. 346), ELLIOTT BROTHERS (LONDON) LIMITED Elstree Way, Borehamwood, Herts.

Due to further expansion, ABC Television invites applications for the following vacancies:-


ABC

## OPERATIONAL CREW

- Sound Balancer - Sound Assistants Grade B

Applicants should have previous TV experience and the ability to carry out first line maintenance.

## ENGINEERING

- Assistant Engineers - Videotape, Telecine, or Communications

A thorough basic knowledge of electronics and the ability to maintain equipment is essential. Industrial experience in an allied industry would be an advantage.

## SOUND DEVELOPMENT ENGINEER

Previous experience in the field of audio development work essential.

## TECHNICAL TRAINEES

Applicants should be over eighteen years of age and hold G.C.E. in at least three subjects, one being Maths or Physics. A basic working knowledge of Radio or Photography would be an advantage.
The Studios are situated in pleasant surroundings at Teddington, by the River Thames, with first class restaurant and recreation facilities. A pension scheme is available subject to an initial qualifying period.

Applications must be made in writing to The Personnel Manager
ABC TELEVISION LTD • BROOM ROAD • TEDDINGTON • MIDDX

## A.R.R.L. RADIO AMATEURS HANDBOOK 1961 32/6

Fundamentals of Radio Receiver Servicing. 3rded, by Squire. Post.I/Beginners Guide to Radio. 5th ed. by Camm. Postage 9 d 1961. Postage 6d

Radio T.V. Tube, Diode and Tran sistor Equivalents Manual by sistor Equivalents Manual by World Radio and T.V. Handbook 1961 by Johansen. Postage I/-......... Radio Servicing Instruments by Bradley. Postage 6d
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Offer an opportunity in an out-of-the-ordinary branch of Electronic Engineering to a

## DEVELOPMENT ENGINEER

with at least H.N.C. standard in electrical engineering or electronics. Experience in high-power electronic equipment an advantage.
Interesting work in connection with
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Apply in confidence, giving full details of qualifications and experience to:

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[^0]:    * Squadron Leader, Royal Air Force Technical College, Henlow.

[^1]:    * Bush Radio Ltd.
    † Full title: Synthetic-resin Bonded-paper Sheets for use at Power Frequencies. Copies of British Standards may be obtained from British Standards House, 2 Park Street, London, W.l.

[^2]:    * Wireless World, p. 287, June 1960

[^3]:    $\dagger$ Wireless World, p. 361, July 1960.

[^4]:    士 Wire!ess World, p. 196, April 1960.

[^5]:    * Post Office, London Telecommunications Region.

[^6]:    * Marconi Instruments Lid., St. Albans.

[^7]:    A. F. BULGIN \& CO. LIMITED, BYE-PASS ROAD, EARKING, ESSEX

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