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That's remarkable value for money.

Considerably less than you could pay for the same performance, accuracy, sensitivity and versatility.

Not that we set out to undercut the competition.

All we wanted to do was to produce the best 100MHz general purpose oscilloscope on the market. But because we started from scratch, we were able to use the latest engineering techniques and advanced circuitry (including many i.c.'s). And this meant we could price the 200 very competitively.

The 200 incorporates all the features you would expect in a first-class modular instrument, plus several new ideas.

To find out about these, please telephone Ashley Stokes on 01-450 7811, Extension 113. Or write to him at the address below.

* UK Price quoted does not include VAT



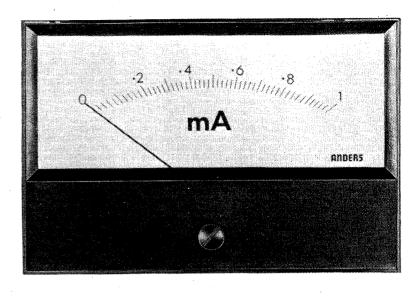
ANDERS MEANS METERS...

REGAL RANGE

- New 100° arc high quality meters at low prices.
- Rugged taut band construction
 pivot and jewel available to order
- \blacksquare Sensitivities to 10 μ A
- Very competitively priced for OEM quantities
- Modern styled meters in matt black plastic cases with flattened arc giving long scale.

TWO MODELS R55 2.5in (63.5mm) Scale length R65 3.2in (81.3mm) Scale length

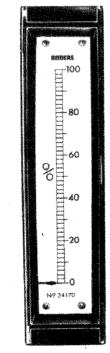
Anders provide what is probably the largest range of meters available from a single source in Europe: MC/MI, dynamometer, vibrating reed, electrostatic, etc. in over 100 case styles and sizes, a few of which are shown below.



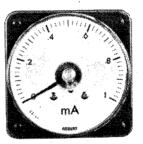
Popular models and ranges are stocked in depth while a specially equipped instrument department enables swift production of non-standard ranges and scales, to suit individual customer requirements, in large or small quantities.



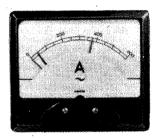
Vulcan Moving Iron. 4 models, 1.5", 1.8", 2.7", 3.7" scales. Voltmeters, ammeters and motor starting meters.



Profile 350 edgewise 4.3" scale.
DC moving coil and AC moving coil rectified.
Horizontal or vertical mounting.



Oxford Long Scale 240°. 2 models, 5.5", 8" scales. DC moving coil and AC moving coil rectified.



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Send for fully illustrated catalogue.

Kestrel Clear Front. 7

scales. DC moving coil,

AC moving coil rectified,

models, 1:3"-5:25"

AC moving iron.

ANDERS ELECTRONICS LIMITED 48/56 Bayham Place, Bayham Street, London, N.W.1. Telephone 01-387 9092.

Manufacturers and distributors of Electrical Measuring Instruments. Sole U.K. distributors of FRAHM Resonant Reed Frequency Meters and Tachometers. Manufacturers of purpose built electrical and electronic equipment to customers requirements.



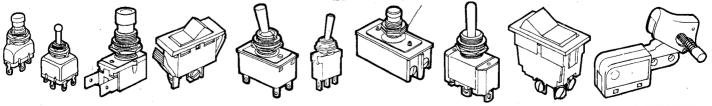
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Arrow switches *right* across the panel



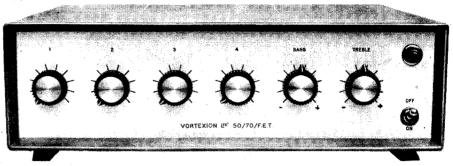
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Vortexion

50/70 WATT ALL SILICON AMPLIFIER WITH BUILT-IN 4-WAY MIXER USING F.E.T.s.

This is a high fidelity amplifier (0.3% intermodulation distortion) using the circuit of our 100% reliable 100 Watt Amplifier with its elaborate protection against short and overload, etc. To this is allied our latest development of F.E.T. Mixer Amplifier, again fully protected against overload and completely free from radio breakthrough.



The mixer is arranged for $2-30/60\Omega$ balanced line microphones, 1-HiZ gram input and 1-auxiliary input followed by bass and treble controls. 100 volt balanced line output or $5/15\Omega$ and 100 volt line.

50/70 WATT ALL SILICON AMPLIFIER WITH BUILT-IN 5-WAY MIXER USING F.E.T.s. This is similar to the 4-way version but with 5 inputs and bass cut controls on each of the three low impedance balanced line microphone stages, and a high impedance (10 meg) gram stage with bass and treble controls plus the usual line or tape input. All the input stages are protected against overload by back to back low self capacity diodes and all use F.E.T's for low noise, low intermodulation distortion and freedom from radio breakthrough. A voltage stabilised supply is used for the pre-amplifiers making it independent of mains supply fluctuations and other stabilised supply for the driver stages is arranged to cut off when the output is overloaded or over temperature. The output is 75% efficient and 100V balanced line or $8/16\Omega$ output are selected by means of a rear panel switch which has a locking plate indicating the output impedance selected.

The mixer section has an additional emitter follower output for driving a slave amplifier, phones or tape recorder, output .3V out on 600 ohms upwards.

100 WATT ALL SILICON AMPLIFIER. A high quality amplifier with 8 ohms—15 ohms or 100 volt line output for A.C. Mains. Protection is given for short and open circuit output over driving and over temperature. Input 0.4 V on 100K ohms.

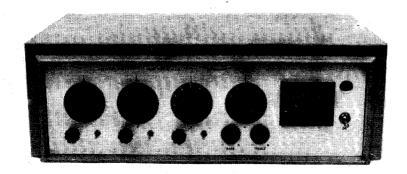
THE 100 WATT MIXER AMPLIFIER with specification as above is here combined with a 4 channel F.E.T. mixer, $2-30/60\Omega$ balanced microphone inputs, 1-HiZ gram input and 1-auxiliary input with tone controls and mounted in a standard robust stove enamelled steel case. A stabilised voltage supply feeds the tone controls and pre amps, compensating for a mains voltage drop of over 25% and the output transistor biasing compensates for a wide range of voltage and temperature. Also available in rack panel form.

CP50 AMPLIFIER. An all silicon transistor 50 watt amplifier for mains and 12 volt battery operation, charging its own battery and automatically going to battery if mains fail. Protected inputs, and overload and short circuit protected outputs for 8 ohms-15 ohms and 100 volt line. Bass and treble controls fitted.

Models available with 1 gram and 2 low mic. inputs, 1 gram and 3 low mic. inputs or 4 low mic. inputs.

200 WATT AMPLIFIER. Can deliver its full audio power at any frequency in the range of 30 c/s-20 Kc/s ± 1 dB. Less than 0.2% distortion at 1 Kc/s. Can be used to drive mechanical devices for which power is over 120 watt on continuous sine wave. Input 1 mW 600 ohms. Output 100-120 V or 200-240 V. Additional matching transformers for other impedances are available.

F.E.T. MIXERS and PPMs.



Since we have been supplying professional mixers for 25 years we have delayed the introduction of solid state units until they were at least as good as their valve counterparts. (Which will continue where required.)

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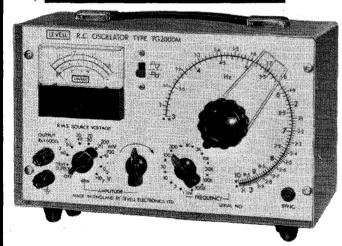
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ANALOGUE

FREQUENCY

1 Hz to 1 MHz in 12 ranges. Acc. \pm 2%

SINE OUTPUT

 \pm 0.03Hz. 7V r.m.s. down to <200μV with Rs

DISTORTION

 $=600 \Omega$

<0.1% to 5V, <0.2% at 7V from 10Hz to 100kHz.

SQUARE OUTPUT

7V peak down to <200µV. Rise time

<150nS

SYNC. OUTPUT SYNC. INPUT

> 1V r.m.s. sine in phase with output.

 \pm 1% freq. lock range per volt r.m.s.

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input. 0/2V, 0/7V & --14/+6dBm. on

TG200M & DM only.

SIZE & WEIGHT

7" high x $10\frac{1}{4}$ " x $5\frac{1}{2}$ " deep. 10 lbs.

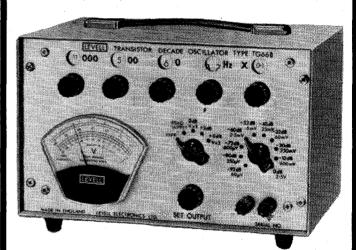
TG200

TG200D

TG200M TG200DM

Sine O/P Sine & Sq.O/P +meter. + meter.

NOTE: All prices subject to V.A.T.



DIGITAL

FREQUENCY ACCURACY

0.2Hz to 1.22MHz on four decade

controls.

 \pm 0.02Hz below 6Hz

 \pm 0.3% from 6Hz to 100kHz

 \pm 1% from 100 kHz to 300 kHz \pm 3% above 300 kHz.

SINE OUTPUT

5V r.m.s. down to 30 μ V with Rs=600 Ω

<0.15% from 15Hz to 15 kHz.

<0.5% at 1.5Hz and 150kHz.

METER SCALES SIZE & WEIGHT

DISTORTION

2 Expanded voltage & -2/+4dBm. 7" high x $10\frac{1}{4}$ " wide x 7" deep. 12 lbs.

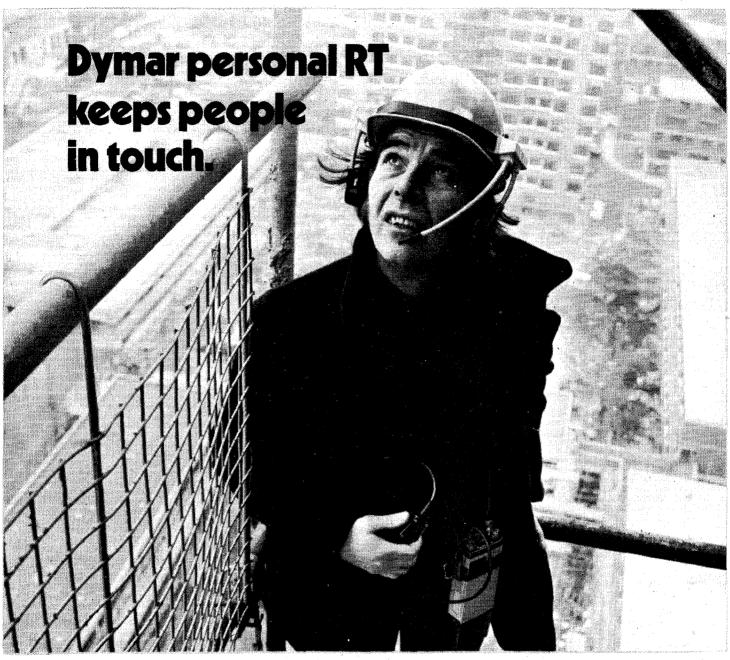
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For more information, use the Reader Enquiry Service, or contact Dymar direct.

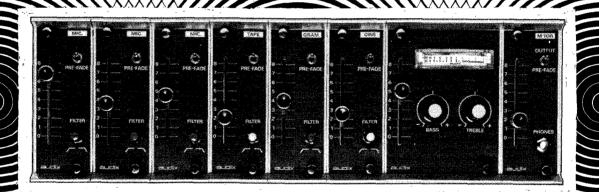


the name in radiotelephones

Dymar Electronics Limited Colonial Way Radlett Road Watford Hertfordshire WD2 4LA. Tel: Watford 21297. Telex: 923035, Cables: Dymar Watford



NODULAR AUDIO MILES



MODEL MXT-200

The MXT-200 is the latest addition to the range of modular audio mixing units available from Audix for public address, theatre, broadcast and recording studio applications. Designed for either mono or stereo working the MXT-200 incorporates high and low frequency filtering per channel as well as overall treble and bass tone controls. A pre-fade listen miniature toggle switch is fitted to all plug-in channels; the maximum number of inputs being 16. The wide choice of modules including output routing, monitoring facilities, P.P.M. or V.U. metering units can be fitted within rack mounting or free standing cabinets.



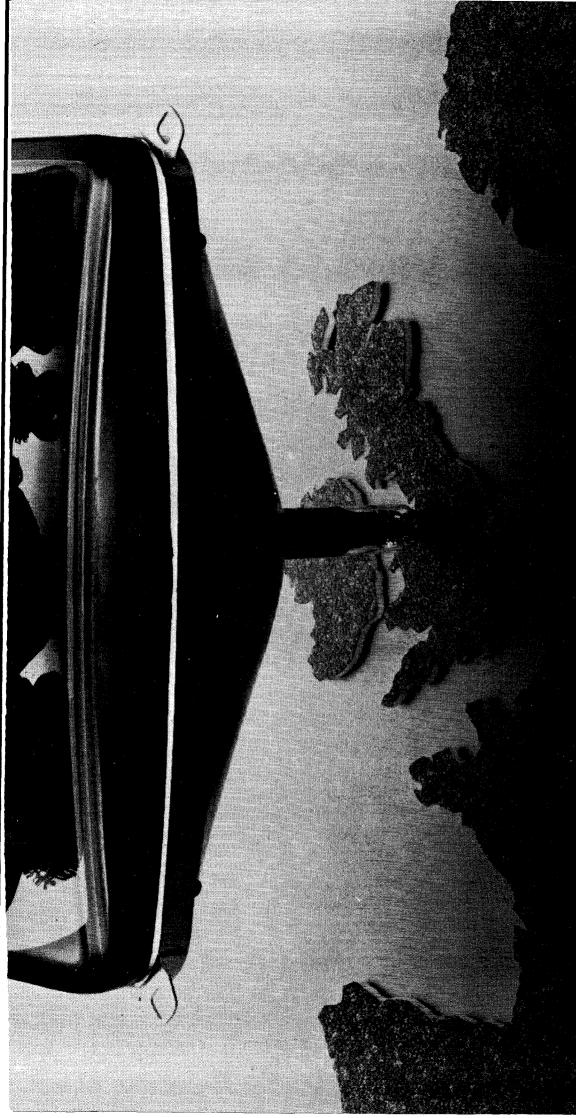
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Which should keep everyone happy – from Wick to Wellington.



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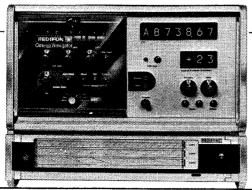
Thorn Colour Tubes Ltd.,

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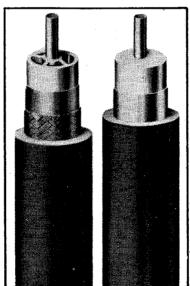
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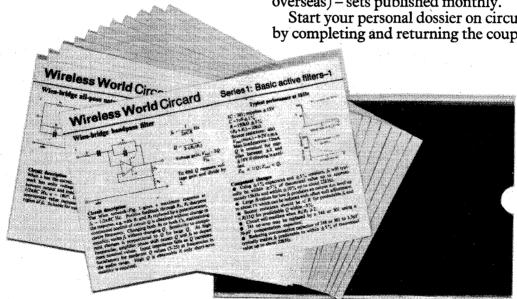
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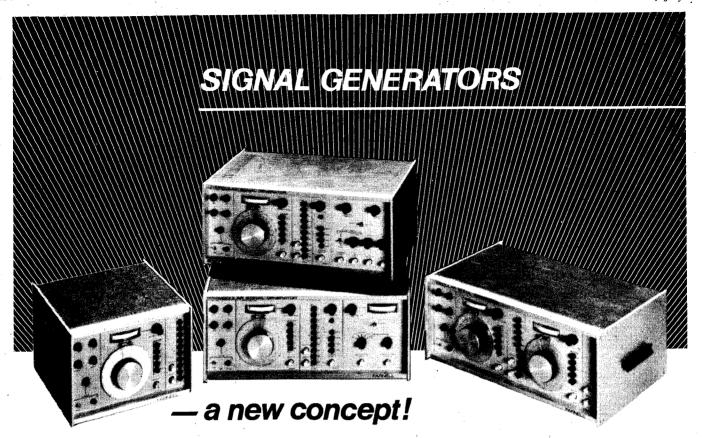
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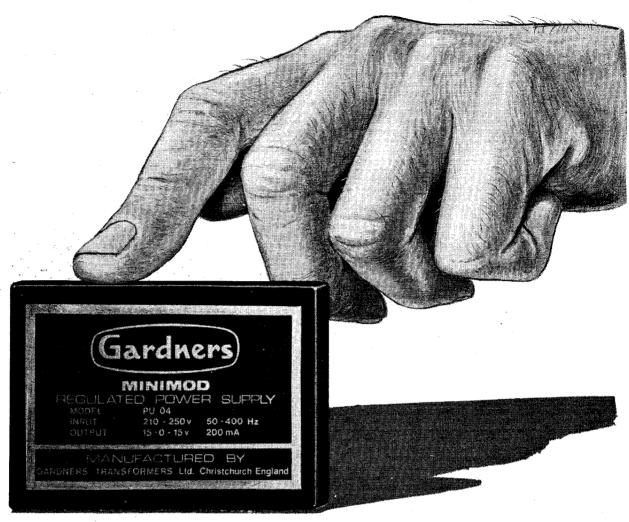
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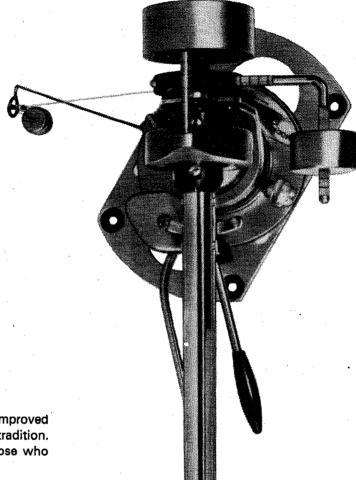
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WW-017 FOR FURTHER DETAILS



3009 v 3009/S.2

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Intended for the best modern cartridges the standard Model 3009 has an effective mass of only 5 5 grams, offering significant improvement in trackability and all round definition plus cleaner bass and low sensitivity to external shock and feedback.

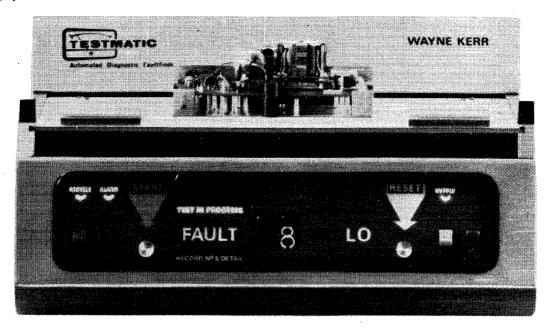
These are the things you will notice when you listen to your records but if a detachable shell means more in your life, use the Model 3009/S.2, it's the next best but its effective mass is 4 grams higher.

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Telephone: Steyning (0903) 814321



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The TM60 tests circuit boards, sub-assemblies, cableforms, and cash flow

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But the TM 60 can be worked by assembly staff and this results in early fault location. By cutting the time between an error and subsequent rejection, productivity is increased and cash flow improved. The modest cost of the Testmatic can be fully recovered in the first few weeks of use.

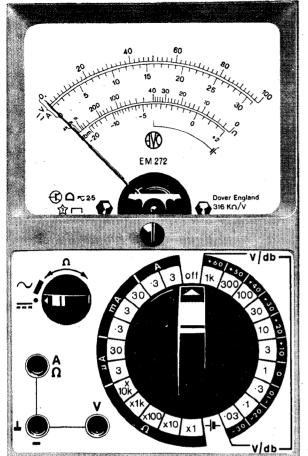
WAYNE KERR

A member of the Wilmot Breeden group

For all the technical data, and further information on the TM 60, please fill in the coupon below. It could well be the first step in saving your company enormous sums of money.

Or quicker still, call us on Bognor Regis (02433) 4501.

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| Company Nameand Address | , |
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Its circuitry achieves maximum reliability by combining printed circuit shunts with thick film modules.

And it has a robust centre-pole movement that takes knocks you'd hate to give it.

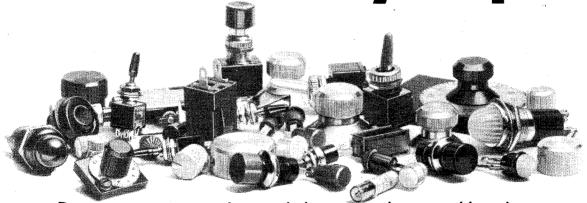
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Avo Limited, Dover, Kent. Telephone: Dover (0304) 202620 Thorn Measurement, Control and Automation Division

WW-020 FOR FURTHER DETAILS

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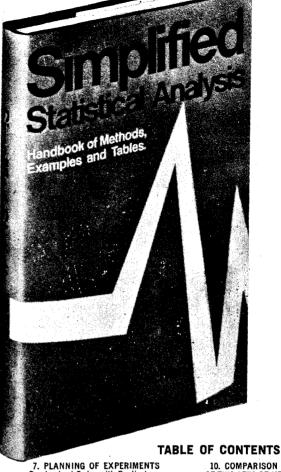
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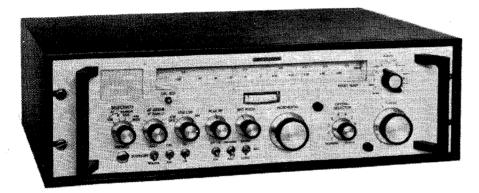
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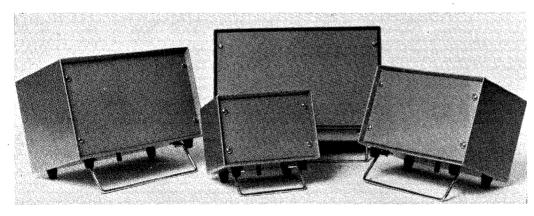
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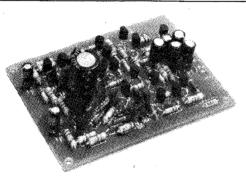
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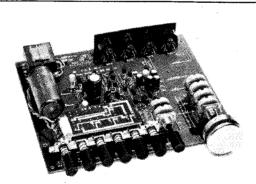
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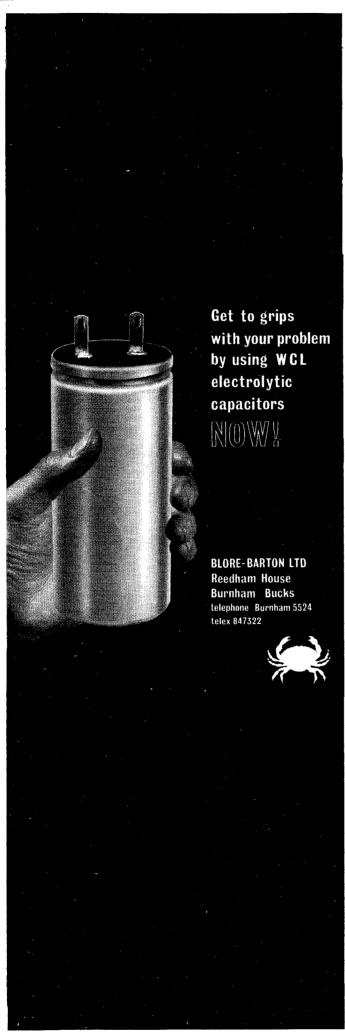
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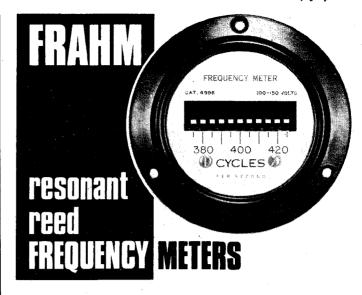
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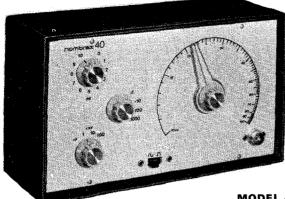
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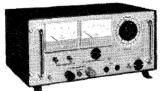
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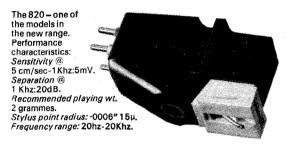
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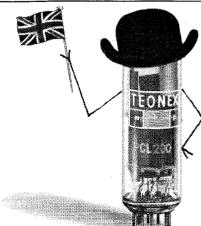
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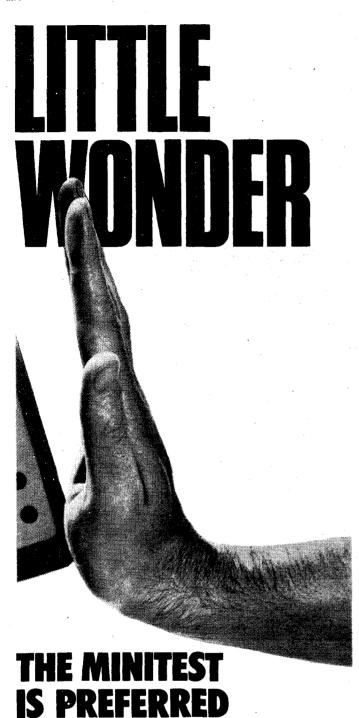
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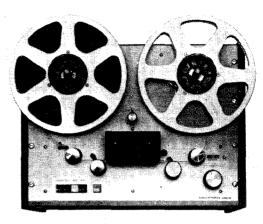
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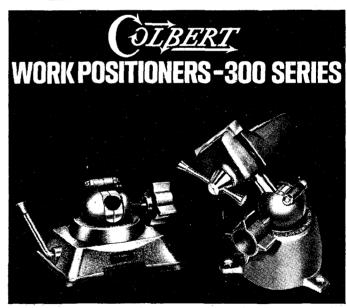
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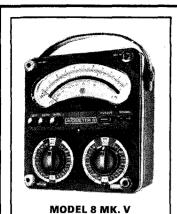
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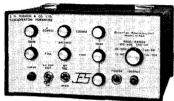
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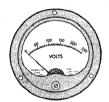
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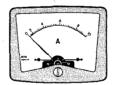
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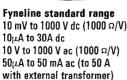
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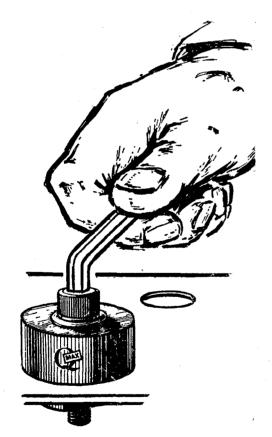
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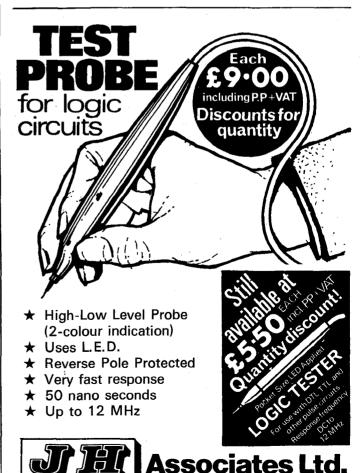
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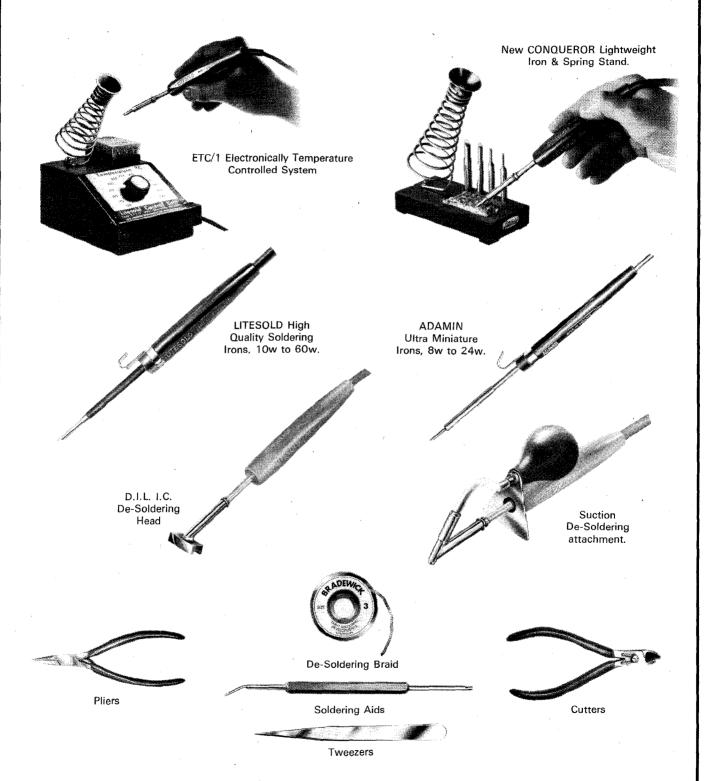
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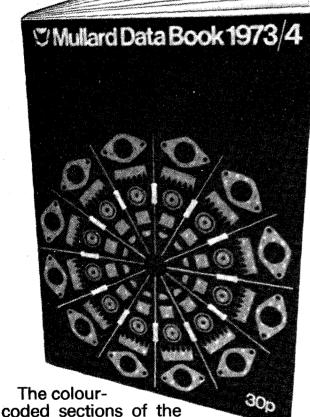


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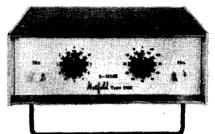
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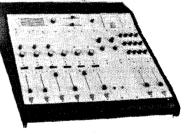
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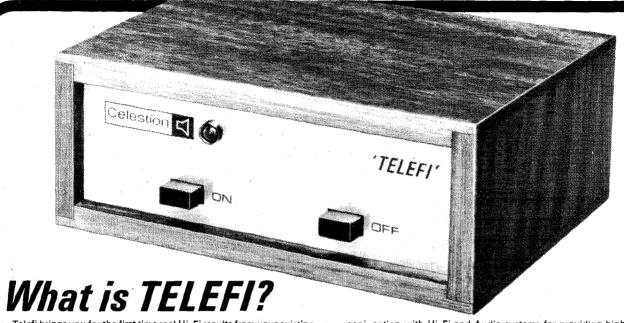
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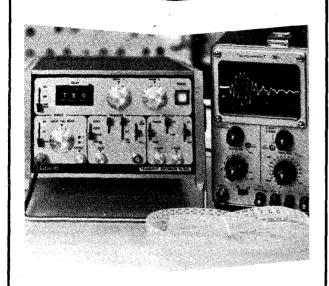
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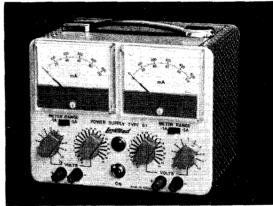
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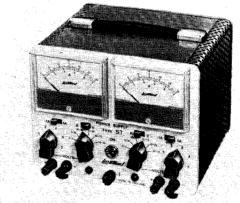
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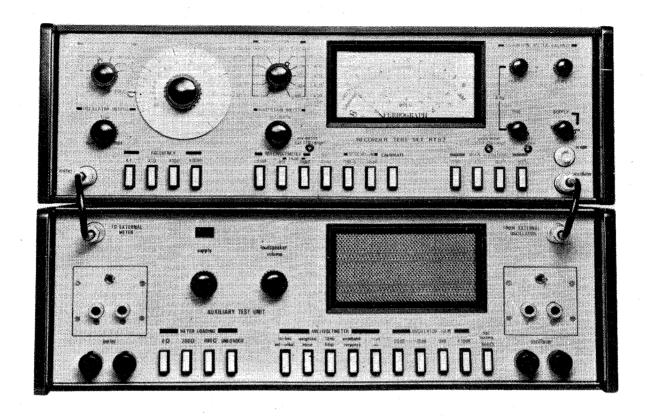
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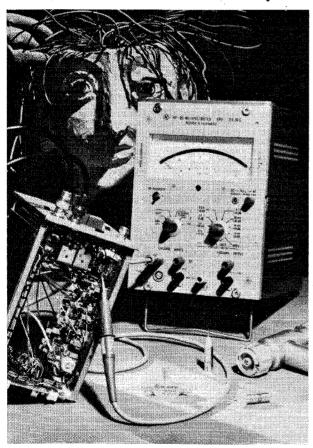
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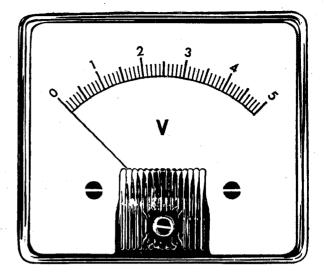
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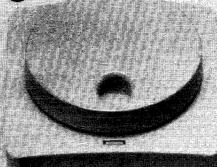
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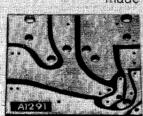
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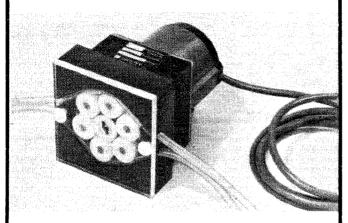
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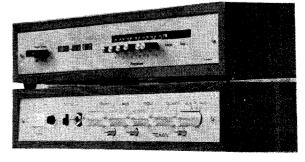
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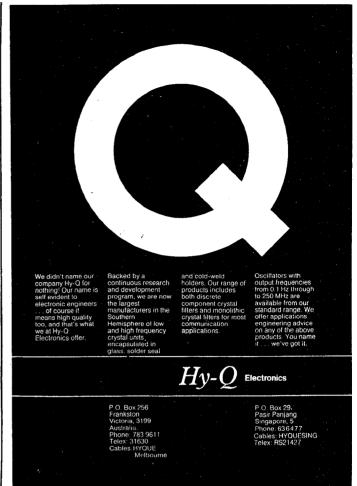


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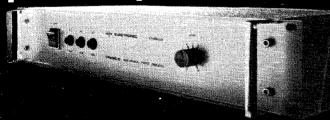


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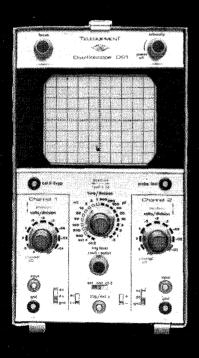
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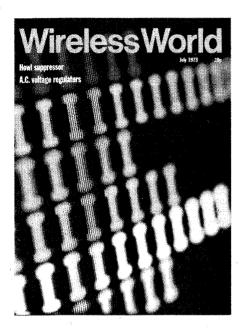
Wireless World

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July 1973

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(Photographer Paul Brierley)

In our next issue

(publication date July 16)

Electronic sound synthesizer. First part of a series describing the circuitry and modular construction of a synthesizer capable of generating a large range of musical and special effects.

Audio amplifiers reappraised. Modern design techniques analysed against the background of existing research and present theories of psychoacoustics.

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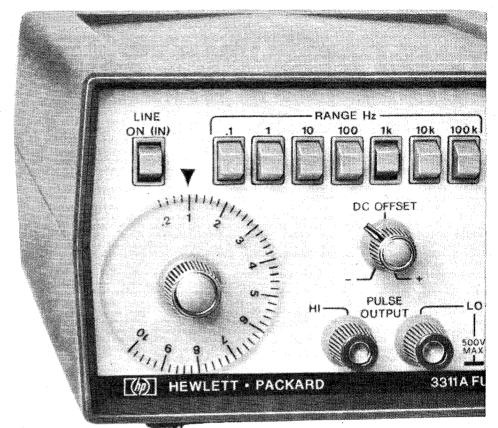
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Wireless World

Choice of Systems and Standards

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The evolutionary process of natural selection, or "survival of the fittest", works reasonably well in determining the design of electronic equipment. Designs which do not suit the customers' needs are not bought and so eventually disappear from the market. There is one situation, however, where successful survival does not necessarily mean that the best solution to technical problems has been found. This is where variation in the design of electronic equipment means much more than different types of hardware for the customer to choose from but implies competing systems of coding and decoding information for transmission, recording and reproduction. We see this situation at present in "quadraphonic" sound equipment and small

Is it a good principle that the choice of an information system for widespread use should be left to the fortuitous working out of market forces?

There are several reasons why it is not a good principle. First, the biggest investment in a particular system is likely to be held by a large and powerful manufacturing group with a loud voice in the market place. Having committed itself to the system by heavy expenditure on research and development, this group will naturally be determined to recover these costs and exploit the investment by using all its resources in marketing. And this strength is likely to prevail, regardless of whether the system itself is a good one (e.g. in performance, economics, convenience, reliability) or not. Secondly, a commitment to a particular system will often be determined, or strongly influenced by, the current economics of making the hardware. Engineers may have longer term ideas of what is best, but the commercial interests of the manufacturing company dictate that it must go ahead as soon as possible making and selling hardware. Thirdly, when the choice of a system is left to the outcome of entrepreneurial activity there will inevitably be a period of great waste. Some manufacturers, those who have chosen the unlucky (not necessarily wrong) system, will waste resources in making and attempting to sell the associated hardware, and this waste is likely to be on a larger scale than that resulting from an unsuccessful product not associated with a system choice. And those customers who actually buy the unlucky hardware will, of course, have wasted their money on non-compatible and hence prematurely obsolete equipment.

Ideally the choice of an information system should be determined jointly by all interested parties — by potential users as well as manufacturers. Where the users are large organizations such as broadcasting authorities this does happen to a great extent, simply because the customer is powerful, rich and technically well informed. Where the users are ordinary citizens buying domestic electronic equipment it can be fairly said that the customers do not know what they want because they are not in a position to judge the merits of competing systems. The nearest approach to making co-operative decisions on systems is when a group of interested parties, mainly manufacturers, is formed on a national basis, such as the Japanese E.I.A.J. (e.g. for video recording). Such bodies are formed to promote and protect mutual interests, not to reach truly objective decisions. But perhaps this

is the best we can hope for in an imperfect world.

ORACLE — Broadcasting the Written Word

by A. James, * F.I.E.E.

Although public broadcasting is now more than 50 years old, it still consists mostly of the spoken word or of pictures of people speaking, and yet for many purposes the written word is more useful. Written material has the great advantage that the reader can set his own pace, can stop to think, and can re-read the text any number of times, at will.

Several proposals for broadcasting written material have been made in the past. One of them, the B.B.C. Ceefax proposal has already been described in Wireless World¹. The Independent Broadcasting Authority has under development a system for transmitting written material which, like the Ceefax system, transmits the information along with the television pictures. In some other significant respects the system differs from Ceefax, however. Experimental field trials of the IBA system have been taking place in the London area since early April.

"Notes in the margin"

* Independent Broadcasting Authority.

Fig. 1 shows a section of the television signal. On the left are the last few lines of one picture; following it comes the frame synchronizing pulses and then come a

termed "the margin" at the top of the next picture. Finally, on the right are the first lines of the next picture. Towards the end of this "margin" in line 19 and sometimes also in line 20, additional test line signals are nowadays inserted². Additional signals are also sometimes accommodated on line 16 and these signals take the form of a binary code used to transmit data along the network³. The Independent Television network uses such a digital data signal to identify the source of transmissions on the network. For source identification purposes only a short signal is required but the format chosen readily allows expansion so that a message of some 10 letters can be transmitted on each line. The IBA proposes to use these "Notes in the margin" also for the broadcasting of written material and have coined the code word "ORACLE" (Optional Reception of Announcements by Coded Line Electronics) to describe the proposed system.

number of blank lines which might be

What ORACLE does

ORACLE is a system for the assembly, transmission and reception of written material, using the existing television transmission facilities and the ordinary domestic television receiver to display the

text. Apart from the data signal mentioned in the paragraph above which in no way affects normal reception, the television transmission is entirely standard. The receiver, however, requires adaptation, either by means of an independent add-on unit or, more cheaply, by an internal modification. The system now undergoing test allows for the transmission of 50 pages of text, each consisting of 22 lines of 40 characters, and the experimental receiver adapter enables the user to select any one of these pages for display. The text may be displayed superimposed on the television picture, like a subtitle, or the viewer may choose to switch the picture off, so that the text is displayed on a neutral background.

At the sending end, the text of a page is prepared in advance on a computer terminal unit. The various pages are then entered and held in the memory of a small computer which allows the information to be inserted into the television signal in the correct format at the appropriate time. When the page has been completed it can be transmitted, the time required by the present experimental system being approximately $1\frac{1}{2}$ seconds, so that the total time for transmitting all 50 pages, in effect a booklet of written material, takes about one minute. At the receiver the viewer can, if he wishes, hold the text indefinitely rather than allowing it to be up-dated. At present the text is restricted to a standard type-face of some 60 characters but the code used will allow for an extension of this. The text could even incorporate non-typographical signs designed to allow the transmission of drawings or even free-hand writing.

Data code

Text is transmitted in segments of ten characters each, and each character is represented by an eight-bit code (see Fig. 2). The character codes are preceded by codes defining (i) the segment of the page currently being transmitted and (ii) the ORACLE page. The system is entirely compatible with other standardized data systems using line 16. The form of modulation used is known as "complemented element" in which "1" is represented by a transition from picture white to black and "0" by the reverse transition. This form of modulation is

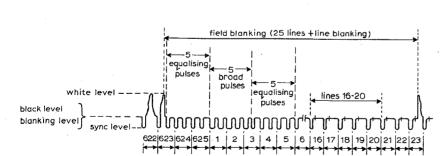


Fig. 1 Part of the television signal showing the vertical interval.

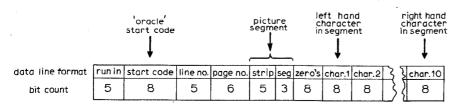


Fig. 2 ORACLE data format for one segment (i.e. one quarter of a line of written information).

particularly rugged and therefore unlikely to suffer from errors during transmission, and has been well tested.

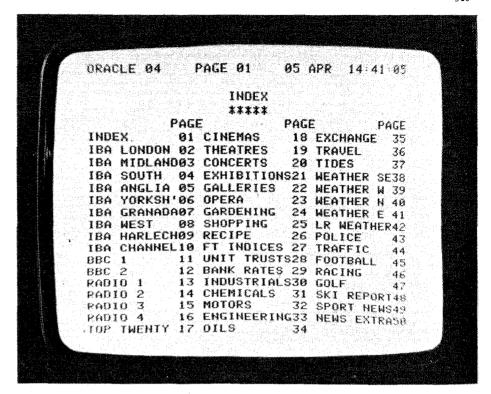
Receiver adapter

The equipment at the receiver (Fig. 3) is in principle simple. The data is extracted from the television signal, and then checked to see whether it is ORACLE data; if it is the page number is read and when this coincides with the page number selected by the viewer, data is fed into a store capable of storing the ten letters transmitted in a single field interval. Subsequently at the appropriate moment, data is fed out of this store into a recirculating store capable of storing one complete ORACLE page of information - that is to say, 880 characters in 88 segments, a total of 5280 bits. A third store is needed to complete the sequence. This store holds a complete line of 40 characters, that is four segments of information, and is updated by the recirculated store. The signal from the third store is presented to the character generator one character at a time.

Each character is based on a matrix of 7×5 dots. On the television pictures the characters are built up line by line by remodulating the scanning beam in the appropriate way, i.e. the scanning beam must be modulated in rapid succession by the first row of squares for each of the letters in a line of text, followed by the second row and so on until the complete line of text has been displayed. To achieve this the information from the third store is presented character by character in rapid succession seven times in all. The line of text is followed by space four television lines wide. Before these are scanned the third store is reloaded from the main recirculating store. Pressing the hold text button isolates the recirculating store from the input store. The data is then recirculated and the same page of text is presented until the button is returned to normal. On switching to a different ORACLE page the recirculating store is emptied and then waits until the first segment of the new page is available. The new segment is then written in to the recirculating store in precisely the same place as the first segment of the old page. Further segments are entered as they become available and in the course of approximately $1\frac{1}{2}$ seconds a complete new page is transferred to the recirculating store.

Possible uses for ORACLE

It is foreseen that the type of information services which could be provided by such a public broadcast data service are almost unlimited, and the system has many other potential applications for home or business use. The type of information services which could be provided in this way includes continuous accurate time checks, weather forecasts, news, traffic and local news, stock exchange prices, television and radio programme information, recipes, local and national announcements, exhibitions, theatres, shopping information and advertising messages. One page of



ORACLE's index page.

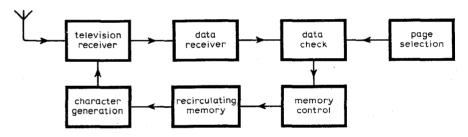
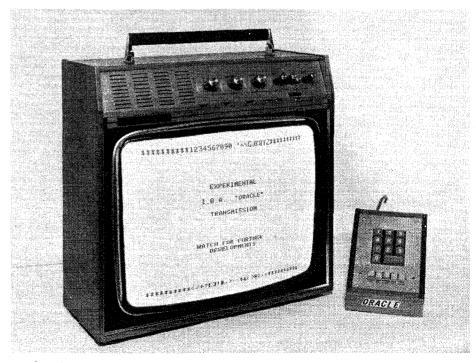


Fig. 3 Block diagram of an ORACLE receiver unit.



Receiver modified for use with the ORACLE system.

information would provide a summary of the services available and the "page number" which would be used by the viewer to "call up" the display on his screen. Other possibilities include the use of the equipment as a home data terminal, the provision of caption material to accompany television programmes for the deaf viewer or for those who do not understand the language of the programme.

At the studios, page information would be kept up-dated by an editorial team using visual display units and keyboards in conjunction with a central computer which would be needed to collate the information so that it can be inserted into the television waveform at the appropriate moments and distributed along with the normal television signals to the transmitting stations. The first line of each page would carry an identification, date and time so that every viewer, no matter which of the services he was watching, would have the equivalent of a highly accurate digital clock always available. For example the first line might read: "ORACLE 33" PAGE 22 21 APR 17: 21: 07. Sufficient flexibility is incorporated in the system contemplate further possibilities such as the use of individually coloured letters or graphical symbols.

The future

The system described above is currently being field-tested in the London area. Broadcasts of test material are being made from time to time and in addition at other times a test caption is being transmitted. The receiver industry is showing great interest in this development and at least one manufacturer has a receiver adapter in an advanced stage of design.

References

1. "TV Information Service", Wireless World, May 1973, p.222.

2. Schaffer, J. and Lever, I., "Insertion Test Signal Equipment", Report of the Joint Conference on Television Measuring Techniques, IERE, May 1970, p.445.

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Applications of the high-standard l.f. Source

by J. M. Osborne*

The uses to which the "High-standard Low-frequency Source" (W.W. January 1973) can be put are many. The source utilizes the B.B.C. 200kHz Droitwich transmission and provides a range of frequencies up to 20kHz. The following are two applications which the author has found particularly useful.

Watch tester

Most watches tick a whole number per second, typically five ticks per second. Their timekeeping can be accurately and quickly tested using the high standard frequency source with the system shown in Fig.1. The watch is placed in contact with a microphone, preferably in a soundproof box. The output of the microphone is amplified and applied to one input of a double beam scope which is set to run at about five sweeps per second, so that a noise pulse from the watch ticking appears near the centre of the trace. A 100Hz square wave from the standard frequency source is applied to the other trace and used to trigger the c.r.o. If the watch gains or loses the pulse will drift across the screen and the time taken to gain or lose one hundredth of a second (one complete cycle of the square wave) is readily determined on some other timepiece. It takes a few minutes to see that a watch is accurate to within a minute a month.

The apparatus can be set up using a cheap watch with a loud tick, which is more likely to show a significant error.

Frequency meter

The l.f. source can readily provide a gating signal for frequency measurement of audio frequencies. The signal to be measured is fed at a suitable level to one input of an SN7400 gate (Fig.2), and the gate output is used to drive a counter. The other input of the SN7400 is connected to the source which is switched to 0.1Hz so that the gate opens for 5 and closes for 5. Thus the counter records the number of cycles of the signal occurring in five seconds. There is a five-second pause to take the reading before the next count. As the counter is cumulative one takes the sum of two counts and divides by ten to obtain the frequency.

As the instrument uses part mechanical counting it can only handle frequencies below about 2kHz. Higher frequencies would have to be divided first to bring them in range, e.g. by the use of SN7490 dividers.

*Westminster School

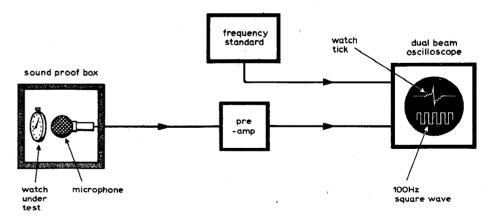


Fig. 1. Schematic of the system for checking watch accuracy.

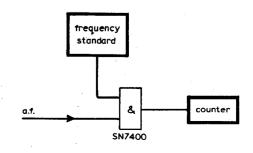


Fig.2. Using the standard in a system for frequency measurement.

Frequency Shifter For "Howl" Suppression

M. Hartley Jones*, M.Sc., Ph.D.

The circuit described increases the stability margin of sound reinforcement systems by up to 8dB. This is achieved by shifting the frequency of signals by 5Hz. The inexpensive and reliable nature of the circuit is made possible by the use of integrated circuits. The technique may be adapted to produce larger frequency shifts for use in electronic music synthesis.

Anyone who is involved with sound reinforcement systems is familiar with acoustic feedback. A sound amplification system where the microphone and speaker are in the same room, is limited in gain by the positive feedback loop formed by the acoustic coupling between loudspeaker and microphone. As the gain is increased, feedback initially causes colorations in the sound, then develops into an audible ringing on transients, finally becoming a continuous "howl-round" when the loop gain reaches unity.

The use of directional microphones and loudspeakers can help in reducing the proportion of sound energy which is fed back to the microphone but, owing to the diffuse nature of the reverberant sound field of a room, acoustic feedback still limits the usable gain of the system.

The reason why feedback usually occurs at one particular frequency is an interesting one. In any sound reinforcement system there will be one frequency at which loop gain is a maximum and this will depend upon the frequency response characteristics of the complete feedback chain, comprising microphone, amplifier, loudspeaker and room, as shown in Fig. 1. If frequency response peaks can be ironed out of every element in the chain, then the mean loop gain can be increased without exciting howlround at a particular frequency.

Compensation for frequency response irregularities in microphones and loud-speakers by means of tone controls and equalizers in the amplifier is a well-established technique and attention to this point can often improve the stability margin of a system by several decibels. The frequency response of the room is, however, a much more complicated problem owing to the multiplicity of resonant modes. Fig. 2 shows the typically irregular frequency

response of a room; only the region just above 500Hz is shown, but the characteristic over the whole audio band is of similar form. The dotted line in Fig. 2 shows the average or mean level of response and it is clear that the peak levels can be more than 10dB above the average. Although it is the average response which determines the subjective sound level in a room, as judged by a listener, it is those 10dB peaks that give rise to howl-round in a sound reinforcement system. Thus, if the peaks could be flattened out, up to 10dB

increase in usable gain could be expected.

There have been attempts at electrical equalization of the major room response peaks¹ but, even if the inverse response of Fig. 2 could be exactly synthesized, the peaks and dips shift with variations in microphone position, with temperature and with size of audience, making exact correction impossible.

Schroeder^{2, 3} has shown that the room frequency response can be effectively smoothed out as far as acoustic feedback is concerned by slightly shifting the frequency

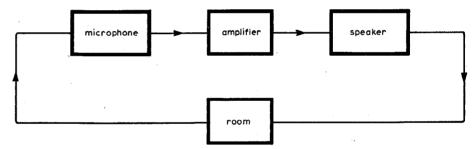


Fig. 1. Acoustic feedback loop in a sound system.

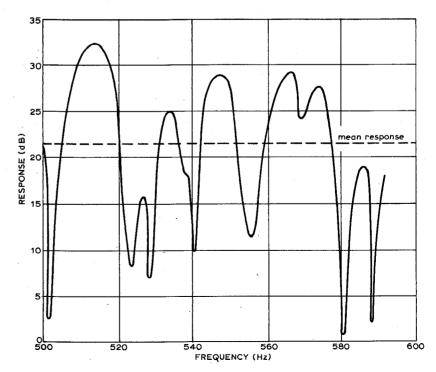


Fig. 2. Small section of a typical frequency response curve of a room.

^{*}University of Manchester Institute of Science and Technology.

of the signal as it goes through the amplifier. This technique ensures that any signal component subject to a very high gain at one of the response peaks, which would normally result in a feedback howl, will be changed in frequency on its next trip round the loop so that it is subject to a much lower gain. As the signal continues to go round the loop it

will be changed in frequency each time and therefore be subject to a different loop gain on each trip. After several trips, the mean loop gain experienced by the signal will thus be equal to the average room response shown in Fig. 2. In the same way, if a signal component happens to experience a low gain on its first trip then it will be subject to

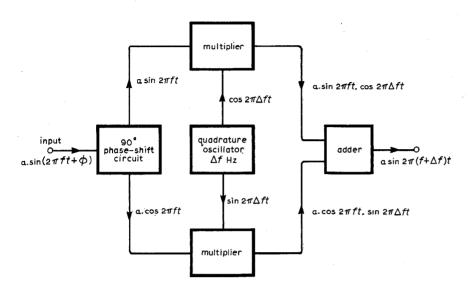


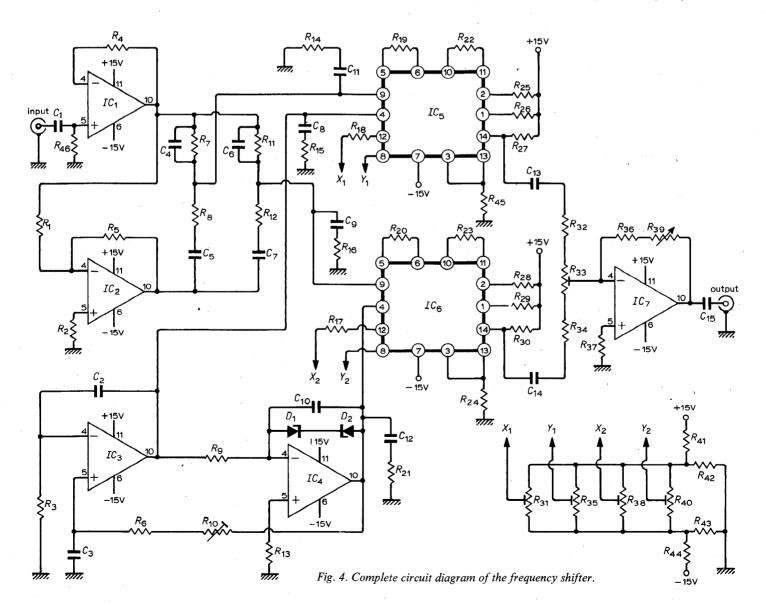
Fig. 3. Block circuit diagram of the frequency shifter.

a much higher gain on its second trip. However, after several trips, the mean loop gain will once again be equal to the average room response. Thus both peaks and dips in the room response are effectively levelled by the frequency shifter.

The amount by which the frequency is shifted is not particularly critical. In order to minimize unnatural effects the shift should be as small as possible, whilst for optimum howl suppression the shift should be of similar order to the average spacing between peaks and their adjacent troughs. Schroeder and Kuttruff⁴ have shown theoretically that the average spacing between adjacent peaks is given by:

$$\Delta f_{\rm max} \approx \frac{4}{T} {\rm Hz}$$

where T is the reverberation time of the room. Thus, on average, in a room with a reverberation time of one second, the spacing between adjacent peaks should be 4Hz and therefore that between a peak and its adjacent trough 2Hz. This average spacing, however, includes tiny irregularities which may be barely measurable; in practice the spacing of major peaks and troughs may be double this figure; in addition, as may be seen in Fig. 2, the peaks are not regularly spaced. The optimum frequency shift for



howl suppression is therefore best found experimentally; in most applications it turns out that a shift of 5Hz is sufficient, no further advantage being gained by greater shifts. It is conceivable that, in the unlikely event of the reverberation time being very much less than one second, a greater shift might be required.

It is immaterial whether the frequency shift is in the upward or downward direction. The shifter to be described produces a 5Hz increase in frequency.

Frequency shift circuit

Schroeder used in his experiments a frequency shift circuit developed by Prestigiacomo and MacLean⁵. This circuit used suppressed carrier single-sideband modulation of a high frequency carrier by the signal followed by demodulation with a new carrier frequency which differed from the original carrier by 5Hz. The circuit to be described here, however, uses a different technique, exploiting the accurate integrated circuit multipliers now freely available; it requires no tuned circuits, crystals or filters and gives a good signal to noise ratio with a wide frequency response and low distortion.

Consider a sinusoidal input signal of frequency fHz; its instantaneous voltage, v_{in} , is represented by

$$v_{in} = a \sin 2\pi f t \tag{1}$$

where a is peak voltage.

Now, if the output signal of our circuit has an increased frequency $(f + \Delta f)Hz$ but the same peak voltage a, then its instantaneous amplitude v_{out} is

$$v_{out} = a \sin 2\pi (f + \Delta f)t$$

= $a \sin (2\pi f t + 2\pi \Delta f t)$ (2)

Expanding this sine function gives:

$$v_{out} = a(\sin 2\pi f t \cos 2\pi \Delta f t)$$

$$+\cos 2\pi ft \sin 2\pi \Delta ft$$
 (3)

Equation 3 is synthesized by the circuit to be described. The cosine functions in the equation are simply sine functions with a $\pi/2$ (90°) phase advance imposed by a phase shift circuit, since $\cos x = \sin(x + (\pi/2))$. Although equations 1 to 3 refer specifically to sinusoidal signals, they may be applied to complex waveforms by Fourier analysis into sine and cosine components.

Fig. 3 shows the basic circuit elements of the frequency shifter. Sinusoidal and cosinusoidal components are derived from the original signal by a broad-band phase shift circuit. An oscillator with quadrature outputs generates sinusoidal and cosinusoidal components of the "shift frequency" Δf . Two analogue multipliers give the products $\sin 2\pi f t \cdot \cos 2\pi \Delta f t$ and $\cos 2\pi f t \cdot \sin 2\pi \Delta f t$ and an operational adder gives the final frequency shifted signal:

$$\sin 2\pi f t \cdot \cos 2\pi \Delta f t + \cos 2\pi f t \cdot \sin 2\pi \Delta f t$$
$$= \sin(2\pi f t + 2\pi \Delta f t).$$

Circuit description

The complete circuit diagram is shown in Fig. 4. The frequency shifter is designed to give unity voltage gain, but an extra 4dB voltage gain is available if required. The

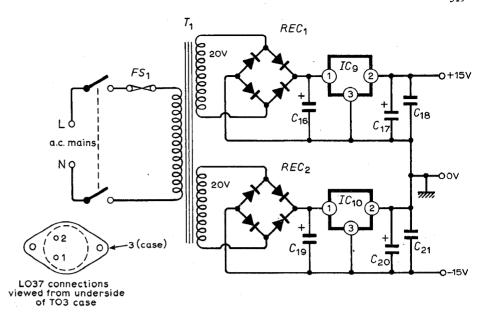


Fig. 5. Power supply circuit.

maximum input voltage before clipping occurs is 4V.r.m.s. (+14dBm, where 0dBm = 0.775V.r.m.s.). The unit is thus suitable for connection between the microphone pre-amplifier and main power amplifier of a sound system. A modification is described later which makes the circuit suitable for use directly in a microphone line.

The input signal feeds first into a voltage follower, IC_1 , giving a high impedance input to the unit. The broad-band phase shifter uses IC_2 as an inverting amplifier and RC networks R_7 , R_8 , C_4 , C_5 and R_{11} , R_{12} , C_6 , C_7 to give two quadrature outputs of equal amplitude. The operation of this circuit is discussed in the appendix.

The sinusoidal and cosinusoidal components of the shift frequency Δf are generated by a quadrature oscillator incorporating amplifiers IC_3 and IC_4 . This operates at 5Hz with the component values shown and gives sinusoidal and cosinusoidal outputs of equal amplitude and with less than 1% distortion. The 6.8V zener diodes across C_{10} act as amplitude limiters, the final adjustment of output amplitude being performed by R_{10} .

Analogue multipliers, IC₅ and IC₆, give the products of the input signal components with the shift frequency components. RC networks C_{11} and R_{14} , C_8 and R_{15} , C_9 and R_{16} , C_{12} and R_{21} are connected directly from the multiplier inputs to earth to suppress any tendency towards h.f. instability. Variable resistors R_{31} , R_{35} , R_{38} and R_{40} are the multiplier input offset controls which are adjusted during the setting up procedure to ensure that when one of the inputs of a multiplier is at zero volts then the output is also zero. Capacitors C_{13} and C_{14} on the multiplier outputs block the d.c. component inherent in the output of this type of multiplier chip and couple the a.c. products to the final stage.

 IC_7 is used in the final stage as an operational adder with R_{39} in the feedback loop to adjust overall gain. Variable resistor R_{33} is a balance control to compensate for any amplitude difference in the products from the two multipliers.

 IC_1 , IC_2 , IC_3 , IC_4 and IC_7 are all type 741 internally compensated operational amplifiers. The Texas Instruments SN72741N 14-pin dual-in-line version was used in the prototype. IC_5 and IC_6 , the analogue multipliers, are type MC1495L (Motorola) or SG1495D (Silicon General). All the pin connection numbers on the circuit diagram refer to 14-pin dual-in-line packages.

Resistors are all $\frac{1}{8}$ W, $\pm 5\%$ tolerance. Capacitor tolerances are all $\pm 20\%$ with the exception of C_4 , C_5 , C_6 and C_7 which should be $\pm 5\%$ or better.

Power supplies

All the integrated circuits in Fig. 4 operate from supplies of $+15\mathrm{V}$ and $-15\mathrm{V}$; the power supply circuit is shown in Fig. 5. Fixed voltage integrated circuit regulators (SGS LO37 or RS Components MVR 15V) are used to produce stable $\pm 15\mathrm{V}$ supplies from a conventional transformer and bridge rectifier arrangement. In the prototype, T_1 was an RS Components 20V miniature transformer and REC_1 and REC_2 were RS Components REC41A. C_{18} and C_{21} are $0.1\mu\mathrm{F}$ polyester or ceramic capacitors to ensure a low supply output impedance at high frequencies.

Construction and adjustments

Circuit layout has not been found critical. The circuit was assembled on a piece of 0.1in matrix Veroboard of size 165×95 mm and housed together with the power supply, which was on a separate matrix board, in a diecast aluminium box of external dimensions $190 \times 120 \times 50$ mm.

In the prototype unit, R_{39} , the gain control, was the only front panel potentiometer. Resistors R_{10} , R_{31} , R_{35} , R_{38} , R_{40} , and R_{33} were all skeleton presets mounted on the circuit board and, once these were correctly adjusted, the settings were found to be extremely stable and required no further adjustment. If no external gain control is required, R_{39} could also be pre-set.

When setting up the circuit, an oscilloscope and sine-wave audio generator are of great assistance but, if necessary, the adjustments can be made using an a.c. voltmeter (e.g. AVO model 8) and a sinusoidal signal from radio, TV or record combined with careful listening. Before switching on it is advisable to set all potentiometers to their mid-position; most of them will then be fairly near the correct setting. The best order of adjustment is as follows:

• The 5Hz sine wave amplitude should be adjusted with R_{10} to 9 volts peak to peak, measured on an oscilloscope at pin 10 of IC_4 . Approximately the same amplitude, but 90° out of phase, should be present at pin 10 of IC_3 . If an oscilloscope is not available for this adjustment, an a.c. voltmeter (1000 Ω /V) may be used, the output being set to 3V.r.m.s. This latter method has proved sufficiently accurate despite some oscillation of the meter pointer due to the low frequency.

• With no signal on the input, R_{31} and R_{38} , the multiplier X-input offset controls, are adjusted for minimum 5Hz component in the output. An oscilloscope or electronic millivoltmeter is essential for precise adjustment; however, if the controls are simply left at their mid-position, the 5Hz rejection will be sufficient for most applications. The response of most power amplifiers is much reduced at such low frequencies.

• With a sinusoidal signal of approximately $1 \,\mathrm{kHz}$ frequency on the circuit input, R_{33} , the adder balance, and then R_{35} and R_{40} , the multiplier Y-input offsets, are adjusted for minimum amplitude modulation in the output, viewed on an oscilloscope. This adjustment may, if necessary, be performed fairly accurately on a listening test using a continuous input, such as a radio or TV tuning signal and adjusting for minimum fluctuation in sound output from a power amplifier and speaker connected to the frequency shifter. A random noise signal (e.g. an f.m. tuner between stations) can also be used for a listening test adjustment.

Although R_{33} , R_{35} and R_{40} do not interact, it may be necessary to go through the adjustment procedure two or three times to obtain optimum results. Once the controls are set, no further adjustment should be required.

Measured prototype performance

In the measurements given below, the overall voltage gain of the frequency shifter was set to unity.

Frequency increase

5.3Hz

Max. input signal prior to clipping 4.0V r.m.s.

Noise level ref. 4V r.m.s. output (measured over a band from 30Hz-10kHz)

-73dB

Residual 5Hz component in output ref. 4V r.m.s.

-66dB

Frequency response

1dB down at 85Hz and 19kHz 3dB down at 45Hz

Residual amplitude modulation (applicable at all output levels)

- <0.5 dB peak to peak output signal fluctuation from 250Hz to 3kHz
- <1dB fluctuation from 220Hz to 13kHz

Total harmonic distortion measured at 1kHz <0.1% at 3.8V r.m.s. output <0.01% at 500mV r.m.s. output

Modification for microphone line

The basic frequency shifter design can be used in any medium or high impedance circuit, but, on a low-level microphone line, noise can become obtrusive. Most of the noise originates in the multipliers; hence the signal to noise ratio can be improved at the expense of overload margin by increasing the gain on the input side and attenuating the output by the same factor.

The simple circuit modification shown in Fig. 6(a) increases the gain of the input amplifier from unity to 27dB by reducing the negative feedback. The voltage divider shown in Fig. 6(b) then provides 27dB attenuation on the output to maintain unity gain overall. After the modification the maximum signal input before clipping is reduced to 170mV, but the noise level is reduced by 27dB. This version of the circuit has given good results in a microphone line with negligible noise increase and adequate overload margin. The unit has also been successfully used in a 30Ω balanced line with the addition of 30Ω to high impedance microphone transformers giving a step up at the input and a step down back to balanced line at the output.

Practical applications

It has been found in practice that the use of the frequency shifter in an average sound system makes available an extra 6 to 8dB of usable gain. Although it was indicated previously that the theoretical gain increase may be over 10dB, this figure is not realized in practice because the feedback that does eventually develop in a frequency shifted system has an unnatural "warbling" character which must be avoided. However, this unusual feedback characteristic has a positive advantage in that a sound system becomes easier to set up and operate because the onset of feedback is so readily detected. When the warbling becomes audible, there is normally an extra 3 to 4dB gain still in hand before instability develops, unlike the unpleasant "ringing" colorations heard in a conventional system which often occur only 1 to 2dB below howl-round.

Even when a conventional system is

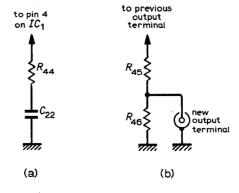


Fig. 6. Circuit modifications for low-level operation: (a) addition to feedback loop of IC_1 ; (b) output attenuator.

operating well below howl-round, the colorations due to acoustic feedback are often sufficient to degrade the sound reproduction so that high quality equipment is unable to give of its best. In these circumstances, the use of the frequency shifter has been found to give a positive improvement in sound quality, even though extra gain may not be required.

It should be emphasized that the circuit obtains its increase in stability margin by effectively flattening the frequency response of the room. In other words, it provides the greatest improvement when the sound from the loudspeaker arriving back at the microphone is of a reverberant character; i.e. when it has been "processed" by multiple reflections within the room. In most systems, even in relatively dead rooms, this is automatically the case; however, if the microphone is only a short distance from the loudspeaker (2–3 metres), then the available gain increase may be reduced to 2 or 3dB.

The frequency shifter has been used successfully not only with speech but also with music amplification. It is perhaps surprising that the 5Hz shift is not obvious, but the only effect noticed has been a slight 5Hz beat between direct and amplified sound on an occasional long continuous note. The general reduction of coloration in the amplification system with the frequency shift in operation outweighed any possible degradation of quality due to beating effects.

Electronic music

Although the circuit as described produces a fixed increase in frequency of 5Hz, the frequency of the modulating quadrature oscillator can be increased by reducing time constants R_3C_2 , R_6C_3 and R_9C_{10} to produce shifts of several hundred hertz or more. As can be imagined, this shifting of a whole band of frequencies introduces weird effects which can be applied to electronic music; all normal harmonic relationships are destroyed because every frequency is shifted by a fixed number of hertz rather than being subject to a given fractional change.

Bode and Moog⁶ have recently described a frequency shifter using the same principle as described here, but specifically designed for electronic music; the modulating frequency is generated by a voltage-controlled beat-frequency oscillator giving a continuous range of frequency shift from –5kHz to +5kHz. By using a subtraction circuit as well as the adder at the output, an upward shift and downward shift may be obtained simultaneously from separate outputs for unusual stereophonic effects. There is obvious scope for the modification of the present circuit along these lines for experiments in electronic music synthesis.

Conclusion

The use of a 5Hz frequency shift in a sound reinforcement system can dramatically reduce acoustic feedback, improving stability margin by 6 to 8dB. Although the technique has been known for over ten years, it has not yet achieved the widespread use it deserves. The circuit described gives high quality results and is inexpensive to build; a simple modification can provide large

shifts of frequency applicable to special effects in electronic music synthesis. The total cost of the prototype was under £20.

Appendix

Design procedure of the broad-band phase shifter. The frequency shifter is a specialized example of the type of single-sideband modulator which depends for its operation on a broad-band phase shifter producing two components of the audio signal with a 90° phase difference. Such phase shifters are also appearing on the domestic scene in quadraphonic decoders; the design criteria of such circuits are therefore of some interest.

The principles of broad-band phase shifter design have been discussed by Dome⁷ and Luck⁸ and analyzed in detail by Orchard⁹. Stein¹⁰ has published calculations on an *RC* lattice phase shifting network similar to the type used here.

To obtain two outputs with a constant phase difference over a wide frequency range, the principle used is to derive the

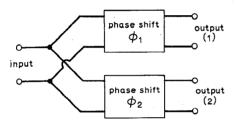


Fig. 7. Block diagram of a broad-band phase shifter.

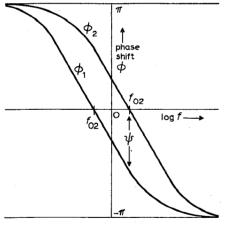


Fig. 8. Graph of phase shifts of individual branches of broad-band phase shifter plotted against frequency.

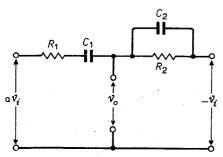


Fig. 9. One branch of the phase-shift network used in the frequency shifter.

outputs from two separate networks each of which produces a frequency-dependent phase shift but where the phase difference between the outputs is constant over the required range. Fig. 7 shows such a circuit in block form and Fig. 8 the type of phase/frequency relationship that will give the required result. It is desirable that the output amplitude from each network should be independent of frequency, i.e. that the network be of the all pass type.

Dome⁷ has shown that a constant phase difference may be realized if the phase shift angle of each network varies according to the logarithm of frequency.

i.e. if $\phi_1 = C_1 + \log K_1 f$ and $\phi_2 = C_2 + \log K_2 f$

where C_1 , C_2 , K_1 , K_2 are constants and ϕ_1 and ϕ_2 the phase shifts of the two networks, then phase difference $\psi = \phi_2 - \phi_1$

$$= C_2 - C_1 + \log \frac{K_2}{K_1}$$

= constant

Luck⁸ has indicated that the desired characteristic may be obtained with a pair of resonant circuits. In the phase angle graphs of Fig. 8, f_{o1} and f_{o2} are the resonant frequencies of the two networks. The required Q-value is very low (<0.5); RC networks can therefore be used. Each network used here is of the Wien type shown in Fig. 9. The following calculation shows that this type of network can fulfil the requirements for a broad-band phase shifter.

In Fig. 9, consideration of circuit impedances gives:

$$\begin{split} \frac{(a+1)v_i}{v_o+v_i} &= \frac{C_2}{C_1} + \frac{R_1}{R_2} + 1 \\ &+ j \bigg(R_1 \omega C_2 - \frac{1}{R_2 \omega C_1} \bigg) \quad (1) \\ &= \alpha + j \beta \\ \text{giving} \quad \frac{v_o}{v_i} &= \frac{a+1-(\alpha+j\beta)}{\alpha+j\beta} \end{split}$$

Now if constant a is chosen so that:

$$a = 2\alpha - 1$$
then
$$\frac{v_o}{v_i} = \frac{\alpha - j\beta}{\alpha + j\beta}$$

$$\therefore \sum \frac{v_o}{v_1} = \sqrt{\frac{\alpha^2 + \beta^2}{\alpha^2 + \beta^2}}$$

$$= 1$$
(2)

Thus the network is of the all-pass type. Now, when $\omega = \frac{1}{R_1C_1} = \frac{1}{R_2C_2}$,

 $\beta = 0$ and the circuit exhibits resonance.

Resonant frequency,
$$f_o = \frac{1}{2\pi R_1 C_1}$$

$$= \frac{1}{2\pi R_2 C_2} \quad (3)$$

It can now be shown that the relationship between phase angle and frequency is of the type required for a broad-band phase shifter.

Substitution for α and β in (2) together

with f_0 from (3) gives:

$$\frac{v_o}{v_i} = \frac{1 + jQ\left(\frac{f_o}{f} - \frac{f}{f_o}\right)}{1 - jQ\left(\frac{f_o}{f} - \frac{f}{f_o}\right)} \tag{4}$$

where
$$Q = \frac{1}{\frac{R_2}{R_1} + 1}$$

and has its usual significance as the quality factor of the resonant circuit. $(Q = f_o/\delta f$, where δf is the half-power bandwidth.)

From (4) the phase angle, ϕ , between v_o and v_i is obtained:

$$\tan\frac{\phi}{2} = Q\left(\frac{f_o}{f} - \frac{f}{f_o}\right) \tag{5}$$

For small values of ϕ , it may be shown¹⁰ that:

$$\phi \approx -4Q \ln \frac{f}{f_o}$$

which is the type of logarithmic relationship required for a broad-band phase shifter. Two of these networks, of resonant frequencies f_{o1} and f_{o2} may therefore be used to produce a constant phase difference,

$$\psi = \phi_2 - \phi_1.$$

Luck⁸ has calculated ψ by applying equation (5) to both networks and has deduced a set of design parameters relating Q, and f_{o1} and f_{o2} with the frequency range over which ψ is to be close to 90°. The greater the permissible deviation from 90°, the wider the frequency range which can be handled.

In the frequency shifter a range from 250Hz to 12kHz is satisfactory and Luck's results indicate that a 90° shift over this range should be possible with maximum deviations of $\pm 5^{\circ}$. The necessary circuit parameters are:

$$f_{o1} = 720$$
Hz
 $f_{o2} = 4,000$ Hz
 $O = 0.21$

thus,

$$\frac{R_2}{R_1} = \frac{C_1}{C_2} = 2.7$$
and $a = 2.5$

The circuit impedances chosen are sufficiently high to avoid excessive loading on the outputs of the drive amplifiers and yet low enough for the $20M\Omega$ input impedance of the multipliers to have a negligible effect on performance. Theoretically the phase shifters should see an infinite load impedance.

Fig. 10 shows the measured phase shift performance of the practical circuit. A phase difference of $90^{\circ} \pm 7^{\circ}$ is obtained from 200Hz to 15kHz.

Deviations from 90° in the phase-shift circuit produce amplitude modulation in the output of the frequency shifter; this occurs at twice the shift frequency owing to the generation of an unwanted sideband. The effect only becomes significant below 200Hz and above 15kHz and is not sub-

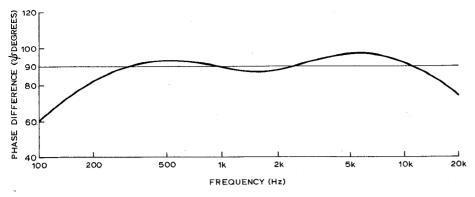


Fig. 10. Graph of measured phase difference plotted against frequency for broad-band phase shift network used in frequency shifter.

jectively troublesome in practice. A closer approximation to a 90° phase difference over a wider frequency range may be obtained if necessary by replacing each Wien network by several such networks in cascade with staggered resonant frequencies.

Components list

| Resistors | |
|---------------------------------------|------------------------------|
| R_{t} | $33k\Omega$ |
| R_2 | $22k\Omega$ |
| R_3, R_4, R_6, R_{46} | $270 \mathrm{k}\Omega$ |
| R_{5}, R_{12} | $82k\Omega$ |
| R_7 | $43k\Omega$ |
| $R_8^{'}$ | 15kΩ |
| R_9, R_{13} | 330kΩ |
| R_{10}, R_{33} | 100 k Ω preset pot. |
| R_{11}^{10}, R_{17}^{33} | 220 k Ω |
| $R_{14}, R_{15}, R_{16}, R_{21}$ | 470Ω |
| R_{18} | $39k\Omega$ |
| $R_{19}^{10}, R_{20}, R_{22}, R_{23}$ | $8.2k\Omega$ |
| R_{24} | 6.8 k Ω |
| R_{25-30}^{27} | $3.3k\Omega$ |
| $R_{31}, R_{35}, R_{38}, R_{40}$ | $10k\Omega$ preset pot. |
| R_{32}, R_{34} | $100 \mathrm{k}\Omega$ |
| R_{36} | 560kΩ |
| R_{37}^{30} | $1M\Omega$ |
| R_{39} | 500k preset or panel |
| | mounting pot. |
| R_{41}, R_{44} | 10 k Ω |
| R_{42}^{41}, R_{43}^{43} | 3.9k |
| | |

Camanitara

Transformer

| Capacitors | |
|--|------------------|
| C_1 , C_2 , C_3 , C_{10} , C_{13} , C_{14} | 100nF |
| C_4, C_6 | 1nF |
| C_5, C_7 | 2.7nF |
| C_8, C_9, C_{11}, C_{12} | 10pF |
| C_{15} | 470nF |
| C_{16}^{13}, C_{19} | $1000 \mu F 40V$ |
| | electrolytic |
| C_{17}, C_{20} | 10μF 16V |
| -179 20 | electrolytic |
| C_{18}, C_{21} | 100nF |
| Tolerance $\pm 5\%$ or better | |

All resistors $\frac{1}{8}W \pm 5\%$ tolerance

| Integrated circuits, Semiconductors | | |
|-------------------------------------|-------------------------|--|
| IC_{1-4} | Texas SN72741N | |
| IC_5 , IC_6 | Motorola MC149SL or | |
| | Silicon General SG1495D | |
| IC_9 , IC_{10} | SGS LO37 | |
| D_1, D_2 | 6.8V zener 500mW | |
| REC1, REC2 | REC41A RS Components | |

T1 20V Min. transformer

RS Components

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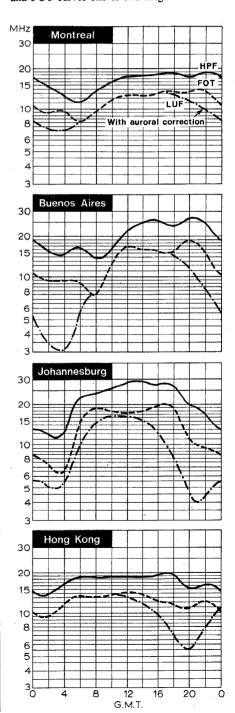
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H.F. Predictions for July

The extended periods of magnetic activity which have been observed each month so far this year should now begin to abate. During these periods maximum usable frequencies reduced and on routes crossing latitudes greater than about forty degrees lowest usable frequencies are increased.

The charts are based on an ionospheric index of 31; this is one third of the value for July of the last year - the effect being to lower HPF and FOT curves one or two megahertz.



News of the Month

Touch terminal communication

A new concept in computer terminal communications — the touch terminal — has been developed by Ferranti and Plessey Radar. The touch terminal comprises a composite display and touch wire system through which unskilled operators may gain access to a computer memory on a question and answer basis.

Touch communication is made possible by the use of touch points incorporated into a mask on the front of a c.r.t. display. When an operator is required to make a decision the computer presents a set of choices on the display, each one being associated with an appropriate touch point. The choices can be displayed as letters, symbols, words or sentences. To make a selection, the operator merely touches the appropriate touch point and a coded signal informs the computer of the choice. The computer then acts on that choice and presents the operator with the next set of choices and so on until the required data is presented (see photo).

Up to thirty two touch points can be fitted to the mask in any layout which suits the customer's requirements. Connections to the touch points are made by a flexible printed circuit which forms part of the mask. Most Ferranti displays

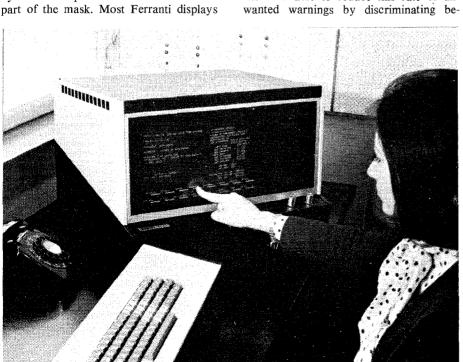
can be supplied as touch terminals and only a minimal amount of re-programming is needed for the computer to use this system. In many cases touch terminals can be used without keyboards as the computer re-annotates the touch points for each set of selections, thus providing the effect of an almost limitless keyboard on the screen of the display.

Communication using the touch terminal is valuable in areas where untrained operators with no typing experience have to communicate quickly and without error with a computer. Typical user areas are information retrieval, process control, air traffic control, reservation systems, simulation, education and many other military and commercial applications.

information retrieval, process control, air traffic control, reservation systems, simulation, education and many other military and commercial applications.

Sonar intruder detector

Security systems based on conventional sonar devices for detecting movement of intruders in premises, have often suffered from spurious alarms caused by inanimate objects flapping or swinging within their range such as curtains, doors, windows, shutters or signs. To combat this particular problem, AFA-Minerva (EMI) has introduced a sonar detector, the Fidela 3, which is able to reduce this rate of unwanted warnings by discriminating be-



tween different types of motion within its range. Its "memory" is able to ignore these typical wavering movements, which previously have triggered off security systems unnecessarily, but is able to identify the approach of an intruder and to sound the alert. Measuring 11in long, 7in wide and $2\frac{1}{4}$ in deep, the device can be fixed to most types of vertical surface. Different patterns of protective field can be provided to meet specific situations.

The Fidela 3 emits ultrasonic waves, and interrogates these signals on their return after deflection from a solid object along their path. Change in frequency of these sound waves caused by any movement within its field of operation is monitored by the equipment's signal processing circuits. These are able to recognize and ignore signals from wavering motions but are designed to trigger the alarm in the event of an intruder entering the protected area.

The detector's basic range pattern is an egg-shaped zone of protection, nominally measuring 25ft long and 17ft at its widest point. This pattern can be easily adjusted to deflect to the right, left or downwards, and be broadened to about 40ft wide with a corresponding reduction in range length to around 12ft.

Local radio frequency changes

Changes of frequency in the v.h.f. services of some of the B.B.C. local radio stations are being made. These changes are necessary in order to conform to a new frequency plan, which will incorporate both B.B.C. and I.B.A. local radio stations.

The changes to B.B.C. stations are in some cases quite small; the new frequencies in MHz are listed below together with the previous frequencies in brackets, and the dates of the changes.

| Humberside | 96.9 (95.3) | 23 May |
|----------------|-------------|---------|
| Bristol | 95.5 (95.4) | 14 June |
| Leicester | 95.1 (95.2) | 14 June |
| Nottingham | 95.4 (94.8) | 21 June |
| Medway | 96.7 (97.0) | 21 June |
| London | 94.9 (95.3) | 28 June |
| Oxford | 95.2 (95.0) | 28 June |
| Brighton | 95.3 (95.8) | 28 June |
| Stoke-on-Trent | 96.1 (94.6) | 6 July |

PAL tolerances

The B.B.C. and I.B.A. are making it clear that although they strive to ensure that their transmitted colour television signals remain within tolerance at all times, they cannot take into account receivers which have not been designed to take advantage of the PAL system. The United Kingdom adopted the PAL system of colour television as its national standard in 1966 after the most exhaustive testing and international discussion. It was seen to offer important advantages over the American N.T.S.C. system although it differs very little in most respects: it was also preferred to the SECAM system.

The advantages of the PAL system

depend upon the use of circuit techniques in the colour receiver which confer upon the picture an almost complete immunity to various forms of colour distortion. These can arise from unwanted changes in the transmitted signals during their passage from the television studio to the home. British receiver manufacturers have consistently designed and manufactured receivers to take advantage of the immunity of the PAL system to colour errors arising from this cause and the signals transmitted throughout the United Kingdom by both the B.B.C. and the I.B.A. have tolerances based upon the assumption that the receiver specification is in accordance with the established principles of the PAL system.

It is, however, possible to design and manufacture a colour television receiver which does not take advantage of the special qualities of the PAL system. Very often the performance of receivers designed and manufactured in this way is perfectly adequate, but circumstances may arise where they do not give a fully satisfactory colour reproduction although the broadcast signals are still within the accepted tolerances.

Award for TV standards converter

Among the recently presented Royal Television Society Awards for 1973 was the Geoffrey Parr Award made to John Baldwin and an I.B.A. design team in recognition of a notable contribution to television engineering. The award is for a digital field rate TV standards converter, which changes colour television signals from one system to another. The new converter is a small device occupying no more than two six-foot equipment racks as

compared with the seven racks of earlier electronic converters. The new unit uses digital techniques and the standards conversion process is claimed to be virtually distortion-free.

More noise reduction

The Burwen audio noise reduction system was recently introduced to the U.K. at a preview given by the distributors, Feldon Audio. Two products were demonstrated: the "Noise Filter", which works to reduce noise and hum in any recorded signal, producing up to 10dB of noise reduction, and the "Noise Eliminator" which is a complementary device like the Dolby but which claims to give up to 50dB of noise reduction.

First quarter TV and radio deliveries

A new monthly peak of deliveries of colour television receivers to the U.K. market was reached in March, 233,000 sets being delivered to the U.K. trade giving a total for the first quarter of 629,000 compared with 363,000 for 1972, a rise cf 73% according to the latest figures compiled by the British Radio Equipment Manufacturers' Association. Monochrome television deliveries reached 115,000, bringing the total for the first quarter to 397,000, a fall of 17% on 1972 (476,000). Deliveries of radios and radiograms reached 438,000 and 27,000 respectively, giving quarterly totals of 1,373,000 and 90,000, representing respective rises of 5% and 41% over 1972 (1,302,000 and 64,000).

These are estimates of U.K. made and

The U.K. Skylab project shown here at the launching pad is based on the conventional two-door range-roving saloon. Four wheels and all-round vision make this another British astro-first. Orbit is achieved by towing the space-lab to its position in space. The lab is attached by a piece of string to its launching rocket (right of centre) and to minimize cost, the launch pad consists of the man in the middle, who holds the rocket and points it at the sky. Ear muffs are provided to deaden the roar on launch. Seriously though — Independent Television News is now operating the smallest outside broadcast unit of its kind in the country. It is a two-camera and v.t.r. unit, with full colour capability.



imported deliveries to the home market, including those to specialist rental and relay companies.

Component service from U.S.A.

A new service designed to obtain U.S. electronic parts for overseas manufacturers more quickly and with less difficulty and expense is offered by Managerial Research, a U.S. consulting firm. Many companies purchase parts in the U.S. but the distances over which communication is required make it costly and difficult to follow up on orders, make substitutions, and arrange for shipments. Managerial Research, working from parts lists supplied by its overseas clients, takes whatever steps are necessary to obtain the needed parts. They are then sent to the client in one shipment, if possible, so as to minimize shipping and handling expenses. More information can be obtained from Managerial Research, Box 240, Wheaton, Illinois 60187, U.S.A.

Briefly

Congratulations to the Association of Public Address Engineers, now 25 years old.

Calculator chip prices. General Instrument Microelectronics have reduced the small quantity price of their calculator single chip C500, to £13.7 — a reduction of £18.

Sixty Years Ago

In an article in our July, 1913 issue on crystal detectors, Dr. A. E. H. Tutton referred to the work of Professor Otto Lehmann on liquid crystals, which are only now being exploited as display elements in digital read-outs. "... the act of crystallisation is not merely a question of the closest packing of molecules in a purely mechanical and happy-go-lucky manner, but the result of the operation of definite attractive and repulsive forces. When these are sufficient to cause the molecules to take up positions such that their centres of gravity or those of small groups of molecules occupy the points of a space-lattice, and then to set themselves further so that these grosser units (molecular or polymolecular) become strictly parallel to each other with respect to the configuration of the molecule (as regards the internal arrangements of its atoms), a solid crystal is produced. But in the cases of the extraordinary compounds brought to our notice by Lehmann, most of which are of complicated constitution (long chains of atoms) and of somewhat viscous, oily, or soapy character, the second state is not accomplished, and the grosser units, probably the molecules themselves in these cases, are wobbly about their centres of gravity; the substance, therefore, either forms fluid crystals with rounded edges and mobile form, readily bent or broken, but as readily recovering themselves, or they are drop-like or even unbrokenly liquid in character, and yet exhibit the optical property of double refraction, and afford brilliant colours in polarised light, just like a solid crystal".

Digital Speed Control Servo

by C. W. Ross, * M.I.E.R.E.

A d.c. motor control system is described which uses a retriggerable monostable pulse as a reference source. The period of a tachometer is compared with the monostable pulse in a simple high-gain pulse-width discriminator driving the motor control elements. The servo system is shown applied to a high performance 2in magnetic tape transport where better than 0.1% speed accuracy is required.

Before going on to the circuit description, it is necessary to mention the tachometer used in the design, as this has some bearing on the techniques used in the electronic section of the unit.

Recent advances in flexible-magnet technology contribute to reducing the cost of the electromagnetic a.c. tachometer illustrated in Fig. 1. The rotor is a disc of 0.06 inches thick flexible magnetic sheet, magnetized with 64 poles. The stator is a single-sided printed circuit board bearing the winding pattern. Fig.2 shows the magnetic pattern of the rotor seen by means of a magnetic particle viewer. The rotor was magnetized on a jig referenced to the centre hole, the magnetic pattern being produced by a multiple-pole electromagnet driven by an s.c.r. and capacitor discharge circuit. The rotor disc is cemented to a rigid metal backplate so that the assembly can rotate with the

motor shaft. The stator is positioned approximately 0.02 inches from the rotor and is rigidly fixed to the motor housing.

The output voltage is sinusoidal and the frequency is proportional to velocity. The prototype gave 40mV r.m.s. per 1000 r.p.m., with a source impedance of less than 1 ohm at 1kHz. Amplitude is proportional to frequency, following a 6dB/octave slope, and is affected by spacing between magnet and pickup.

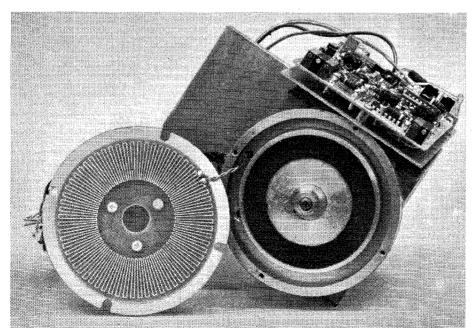
Fig.3 shows the complete assembly attached to the motor. Adjustment is provided to minimize alignment error of the stator by checking the flutter of the complete system with the tachometer frequency conveniently adjusted to 3kHz, the centre carrier frequency for most flutter-measuring instruments. Typically, the flutter is below 0.01% r.m.s. using a 0.5 Hz - 200 Hz bandwidth.

Servo

The trend in speed control systems seems to be toward digital techniques, rather

than the well-established linear methods of operation.

Consider the basic block diagram in Fig. 4. The alternating signal from the tachometer is amplified and amplitude limited by a differential operational amplifier A_1 . The output drives a precision monostable multivibrator that generates a pulse at each positive-going transition of the signal from amplifier A_1 . The complementary output of the monostable circuit is used to drive a pulse-width discriminator circuit, in which the charge and discharge rate of the capacitor is controlled by resistors R_1 and R_2 , respectively. As the tachometer frequency increases, the period of one cycle of tachometer signal will approach that of the monostable circuit. The monostable produces a progressively narrower pulse, because its complementary output is the difference between the uniform "normal" pulse and the period of the tachometer. If the tachometer continues to increase frequency, its period will equal that of the monostable and the complementary pulse will suddenly disappear. The pulse width discriminator circuit senses this rapid change and the charge potential on the capacitor C falls. The potential change across this capacitor is amplified by a high gain d.c. servo amplifier which controls



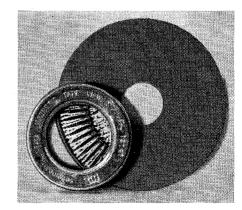


Fig. 2. The magnetic pattern on the rotor, made visible by a particle viewer.

Fig. 1. Tachogenerator, showing the printed-circuit stator winding and the flexible magnetic rotor.

3M Company, Camarillo, California.

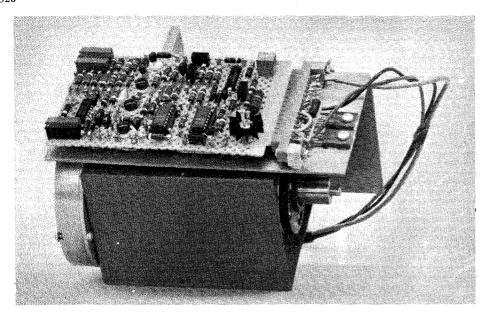


Fig. 3. Complete assembly of motor, tacho and electronic unit.

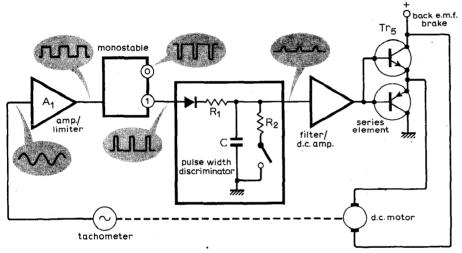


Fig. 4. System block diagram.

the motor. The output of the d.c. power amplifier has complementary transistors connected to enable one transistor to act as a brake; the other is used as the accelerator.

Reviewing this simple system, it can be seen that considerable drift-free gain is available by using the pulse width discriminator driven by the monostable complementary output. Stability and damping are dependent on the RC networks, and the phase lag and/or lead characteristics within the loop.

Electronic design

Tachometer amplifier. A μ A739 dual operational amplifier was used for the tachometer amplifier and d.c. driver amplifier. This is available in a dual in-line package, providing high open-loop gain and differential inputs. An RC network was used to limit the current and voltage output, and to interface it with a 5V logic circuit. The complete servo circuit diagram is shown in Fig.5.

A step-up transformer was used, utilizing the very low impedance of the tachometer to make a larger signal available at the input of the amplifier. The differential inputs are driven push-pull, affording full gain of the i.c. and high common-mode rejection. From very low speeds upwards, the output from the amplifier is a square wave. Only the positive-going edges are used for control purposes, as the particular application for the servo did not warrant use of each zero crossing, which would have added complexity and mark-space adjustment requirements to the design. The design speed range was 600-4800 r.p.m., and half-wave detection (positive-going zero crossings) is adequate.

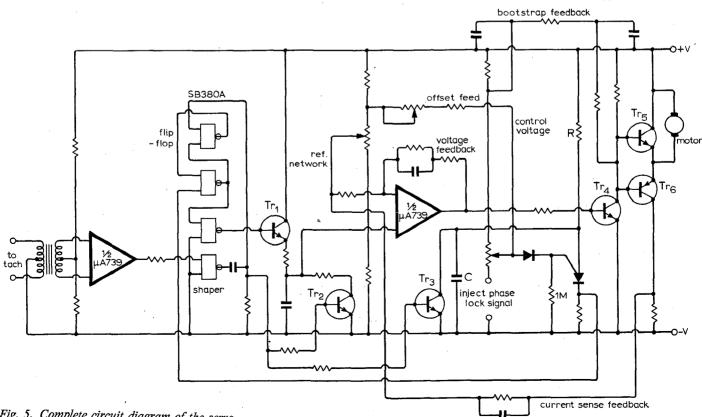


Fig. 5. Complete circuit diagram of the servo.

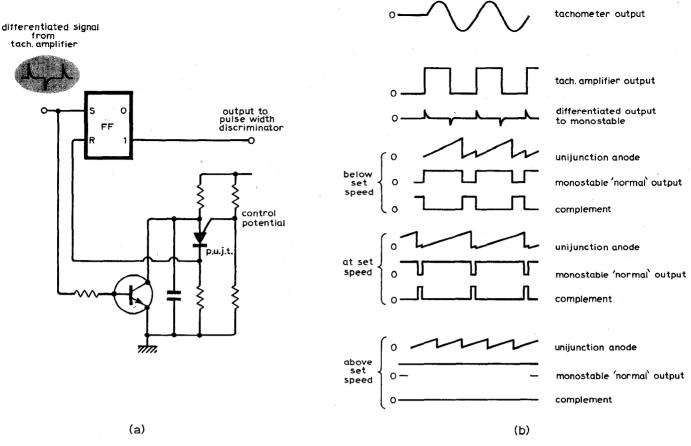


Fig. 6. (a) monostable and control circuit (b) waveforms in the control circuit.

Retriggerable monostable. In the original prototype circuit a 9608 monostable integrated circuit worked well, giving the necessary retriggerable characteristic to enable the 97% or more duty cycle to be achieved. However, temperature stability was found to be a problem when extreme accuracy is required. To overcome temperature stability problems, a lower cost, highly stable circuit was developed. This consists of a quad 2-input gate, interconnected as an R-S flip-flop, and a shaper circuit. The accurate timing pulse is derived from a programmable unijunction transistor, with a switching transistor added to reset the circuit in addition to its self resetting ability when the unijunction fires.

To explain the pulse generation sequence, reference must be made to Fig. 6(a) and (b). Directly upon the application of power, the R-S flip-flop can assume one of its two states. The unijunction timing capacitor charges, and the device avalanches to its negative resistance state, producing a sharp voltage pulse at the cathode resistor. This pulse resets the flip-flop if the set state came up. During this initial period where the tachometer signal is absent, the unijunction operates as a conventional relaxation oscillator. The reset state of the flip-flop is detected as a steady state signal by the pulse width discriminator and the motor accelerates under full power.

The flip-flop is also driven by the positive transitions of the tachometer signal, "setting" the flip-flop at each transition. At speeds below the controlled state, the flip-flop is continuously reset by

the unijunction circuit after a precise delay period (determined by the RC network of the u.j.t. circuit). Notice that the timing capacitor is deliberately discharged by a transistor switch at each positive transition of the tachometer signal; this transistor is driven from the shaper section of the SP380A i.c. The waveforms appearing at various points of the system are shown in Fig.6(b) for three conditions.

The need for the transistor switch to discharge the timing capacitor, in addition to the natural function of the unijunction, will become obvious if the overspeed condition is examined. The capacitor potential is never allowed to reach the peak voltage point of the unijunction at overspeed, and the control voltage to the d.c. power amplifier is zero. This only occurs when an instantaneous command signal is given to slew to a lower speed, or to stop. At control speed, the pulse width from the monostable section of the servo is in the region of a few microseconds to about 15μ s, depending on the stability criteria dictated by the friction, inertia, and compliance of the mechanical loads imposed on the system. The narrow pulses are sufficient to maintain partial charge on the pulse width discriminator capacitor, and the motor power is controlled at the level required to hold the speed selected.

Pulse width discriminator. The pulse width discriminator design is shown in Fig.7. This circuit uses Tr_1 to partially discharge capacitor C after the pulse has charged C to peak value. The discharge transistor is driven via the differentiating network, using the same pulses that drive the

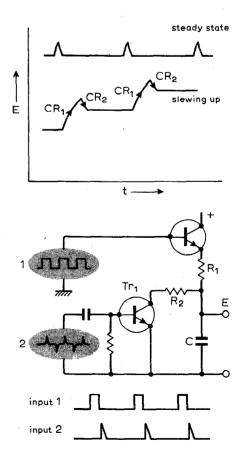


Fig. 7. Pulse-width discriminator, with waveforms.

monostable, hence the time displacement.

The object in this development was to produce a circuit that provided efficient conversion and low, easily filtered ripple content. Ripple can appear as a large portion of the motor current causing a.c. fields around the motor and audible noise at the tachometer frequency, both undesirable when the drive system is used in a magnetic tape transport.

The bandwidth of the servo is dependent on many things: the magnitude of the charge and discharge increments, the pulse width, the value of the capacitor C in Fig.4, and the reference voltage presented to one input of the d.c. amplifier. It was found that this reference voltage required adjustment for each speed, because the average value of capacitor C potential changed with variation of pulse repetition frequency when the optimum pulse width was found. The width of the pulse determines the closed-loop gain of the system, and further d.c. feedback was injected into the reference networks (Fig.5) so that optimum pulse width was obtained over the whole speed range using feedback derived from the main control voltage network. Good tracking was maintained over the required speed range. Measurements made on the pulse width discriminator used in the system illustrated in Fig.5 showed that a 1μ s increment in pulse width produced 0.6 amp d.c. change in the motor circuit, which is ample gain to maintain efficient control.

Power amplifier. The d.c. power amplifier section shown in Fig.5 uses the other half of the μ A739 operational amplifier, driving an output stage. One input of the μ A739 is supplied from a d.c. potential divider; the other input is taken from the pulse width discriminator output filter. The resulting differential is amplified so that a large direct current level is available to energize the motor. Current feedback is obtained by a low value resistor effectively in series with the motor; the resulting voltage is fed back to one input of the preceding µA739 amplifier, reducing the overall gain to a convenient value. Another two feedback paths are used, one being conventional negative feedback and the other a reference tracking voltage derived from the p.u.t. gate control, potential. This offset voltage ensures optimum control pulse width over the complete speed range.

Two output transistors are used, in complementary symmetry configuration. The p-n-p transistor controls the motor drive current, while the n-p-n does the braking when slewing to a lower speed or to a stop, using the back e.m.f. of the motor.

Performance

Accuracy. The speed accuracy was, as expected, proportional to the tachometer frequency and directly related to the tachometer pulse width feeding the peak detection network. To maintain the closed loop gain at a high value, a low impedance supply was necessary to supply the conventional permanent magnet motor

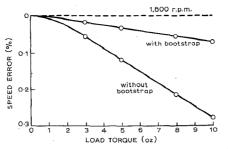


Fig. 8. Speed regulation curve of the system, with and without the bootstrap feedback circuit.

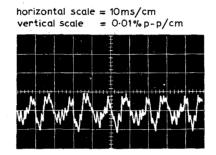


Fig. 9. Residual flutter of the system, showing a fundamental at 50Hz.

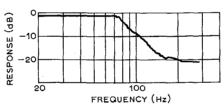


Fig. 10. Bandwidth of the servo only.

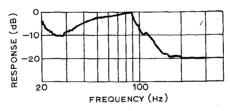


Fig. 11. Bandwidth of the system, showing the effect of compliance and mass of the drive.

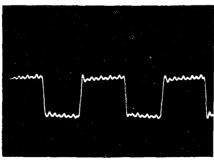


Fig. 12. Response of the system to a 5Hz square wave injected into the control network.

rated at 12 in. oz. continuous torque output.

The speed regulation comparisons are shown in Fig.8 for a midrange speed value. Improvement in speed regulation was achieved by using a d.c. "bootstrap" connection from the collector of the driver transistor Tr_4 in the d.c. amplifier section to the gate control network of the unijunction transistor in Fig. 5. The family of curves shown were obtained by checking speed regulation with and without the "bootstrap" connected.

Bandwidth and transient response. The ability of the servo to follow a sinusoidal speed variation by means of injecting a low frequency signal into the network controlling the p.u.t. timing is a measure of the system bandwidth, or response to a rapidly changing "instruction".

To measure the actual fluctuations of the servo, a flutter meter with a centre carrier frequency of 3000Hz was used to monitor the tachometer signal. The servo speed was adjusted to give a 3000Hz signal and the resulting flutter was measured, first without the sinusoidal programme, then with it. The frequency was varied from a few Hz to over 100Hz until the magnitude of the sinusoidal flutter component reduced by 3 dB. The residual flutter of the system alone is shown in Fig.9, which is typical for this design. The bandwidth graph is shown in Fig.10. Transient response was also measured in a similar way, by injecting a low frequency square wave into the control network. The resulting response waveform was displayed on a cathode ray oscillograph, and is shown in Fig.12 demonstrating the inherent near critical damping of this servo system. The response to a 1% step increment of speed was achieved in less than one revolution. The bandwidth of the servo was such that no ill effects were produced by the natural mechanical resonance of the motor shaft and tachometer motor assembly, as the resonance was above 850Hz.

When the motor was coupled to a typical tape transport using a Mylar belt and flywheel arrangement, a new set of conditions became apparent. The flywheel mass and the belt compliance form a resonant circuit of a significant Q value. Should the "gain" of the servo be high, oscillation can occur within the passband of the system, and adjustment of control pulse width ("gain") is required to give optimum stability. The action is analogous to the electronic a.c. amplifier when there is a resonant circuit within the feedback loop. Provided that the gain and phase shift are present, oscillation will commence (if they are present in excess of the classical requirements for stability). At first sight the belt and flywheel of the servo coupling drive is outside the "velocity loop", but it does feedback heavily, using the motor as a transducer to upset the control ability of the device. This results in the modified bandwidth plot of Fig.11.

Application. The reason for developing this particular servo was to incorporate it

in a precision tape transport, to replace an a.c. two-speed hysteresis synchronous motor. The 3M professional 16-track recorder was used as a particularly good example to evaluate the servo drive, the two-inch tape producing a high load, and the differential "Isoloop" capstan extremely low wow and flutter performance over a bandwidth of 0.5-200Hz. The servo drive was shown to perform well, giving reliable low flutter performance figures of around 0.035% r.m.s. at 15 inches per second. Long-term speed stability and variations due to load change were found to be better than 0.1%.

The main sources of temperature drift are the base-emitter and saturation potential variations in Tr_1 and Tr_2 (Fig.5). The p.u.t. is compensated by the diode in series with the gate. The variations in Tr_1 and Tr_2 are complementary and tend to cancel as do the p.u.t. gate and diode combination. Typical spread on a pilot run of printed circuit units was $\pm 0.2\%$ speed variations for 0°C to 55°C temperature range. The actual environment seen in studio recorders is likely to be + 20°C to + 40°C. Some drift is also present during "warm-up" due to self-heating of the various components of the servo, but these effects were consistently below 0.1%.

The speed could be continuously varied over a 16:1 range, or switched to set values between 300 and 4800 r.p.m. A single resistor value can control the speed. The gate voltage of the programmable u.t. is the parameter which controls the speed - low voltage at this point produces shorter timing pulses from the hybrid monostable circuit, high voltage produces long pulses, and the servo will run at a speed where the pulse duration corresponds to the period of the tachometer. The limiting factor here is the extremes of voltage at the gate cf the p.u.t. — the device has a low limit of a few volts, the upper limit being a potential few volts below the maximum rating for the device itself.

Another important requirement for range of control is the motor rating. At the high end of the range, there must be a good margin of potential left for the servo circuit to use and to provide effective control. The electromagnetic tachometer output must be high enough at the extreme low end of the speed range to produce the correct limiting action from the tachometer amplifier.

The servo could be phase locked to a sinusoidal signal within 2% of the tachometer frequency. The servo was mildly sensitive to amplitude of the synchronizing signal, the optimum point being about 0.5V r.m.s. injected into the lower end of the control network. This limited capture range is inherent in this type of integrated pulse system, allowing only one pulse to be generated by the monostable per period of tachometer, for satisfactory operation.

The author wishes to thank the 3M Company for giving permission to publish this paper, and credit is due to the engineers of the Mincom Division; Jack Mullin, Pete Amass, Ken Clunis, and

others in the Audio Lab for their help and ideas during the development and in the preparation of this paper.

Further reading

Servomechanism Practice, Ahrendt and Savant, McGraw-Hill General Electric Application Note 90.70 Motorola Application Note AN-445

Appendix

| Electrical gain motor-to- tachometer | =64 rad/rad |
|--|-------------------------------|
| Electrical gain motor-to- tape | =2880 rad/rad |
| Discriminator gain | =0.053V/rad |
| Voltage amplifier gain | =16V/volt |
| Power amplifier gain | =17.7A/volt |
| Inertia of motor and tachometer | =0.006 oz in/sec ² |
| Bandwidth, basic servo only | =100 Hz |
| Bandwidth, total drive system | =90 Hz |
| Residual flutter of tachometer | =0.01% r.m.s. |
| Residual flutter from tape | =0.035% r.m.s. |
| Residual carrier leakage current | =0.07A p-p |
| Residual rotational errors | =0.05A p-p |
| No-load direct current | =0.33A |
| Typical operating system current | =0.7A |
| On-off slew rate — up, basic servo | $=4000 \text{rad/sec}^2$ |
| On-off slew rate — down, basic servo | =800 rad/sec ² |
| Incremental slew rate, up | =270 rad/sec ² |
| Incremental slew rate, down | $=113 \text{ rad/sec}^2$ |
| Long-term speed accuracy at 25°C | =0.1% |
| Temperature coefficient | $=0.01\%$ per $^{\circ}$ C |
| Signal-to-noise ratio r.m.s. ripple/max. drive current | =34dB |
| Flutter rejection coefficient open-loop gain | |
| closed-loop gain | =55dB |

Obituary

G. G. Gouriet

The early death of Geoffrey Gouriet at the age of 57 is a great loss, not only to his colleagues in the B.B.C., but also to his many friends and associates.

After various jobs as a clerk, service engineer and design and research engineer, he joined the B.B.C. in 1937 as a junior maintenance engineer. He was soon transferred to the medium-wave transmitting station at Penmon in Anglesey, and there spent all his spare time in studying the fundamental aspects of radio and electronic engineering. Most young people aspiring to the heights of an engineering career need a university training or the equivalent, but Mr Gouriet was quite capable of thinking for himself and this contributed to his great capacity for original work - he was the author of 31 patents and he wrote a considerable number of scientific and engineering papers. In 1939 he transferred to the crystal drive section of the B.B.C.'s transmitter department at the research premises in Balham, where his work had a major influence on the success of the method of operating several transmitters on exactly the same wavelength, thus enabling broadcasting to continue during air raids whilst preventing enemy aircraft from using the broadcast signals for direction finding.

In 1943 Mr Gouriet joined the research department, at first in the aerial section and then in television where he became Head in 1950. From that time until, in 1958, he resigned from the B.B.C. to become technical director of Wayne Kerr Laboratories, he made great contributions to the science of television. The method of equalizing the linear distortions imposed upon a television signal by the apparatus through which it passes, by means of a set of time derivatives of the signal was perhaps his most important technical contribution and although the method had been used in certain specific cases, he clarified and generalized the theory, making it available to all. In 1956 he represented the B.B.C. on the International Radio Consultative Committee of the International Telecommunication Union in the U.S.S.R. and Poland.

At Wayne Kerr he maintained his interest in derivative equalization and applied his ideas to the clarification and improvement of certain aspects of feedback and control theory, particularly as they concerned the rapid but precise movement of heavy mechanical equipment.

In 1964, Mr Gouriet rejoined the B.B.C. as head of research, bringing to this work his special gifts of enthusiasm, friendliness, technical integrity and an intuitive but incisive mind. In 1968 he became chief engineer, Research & Development; &s overall chief of the research and designs departments. He had been elected chairman of the Electronics Division of The Institution of Electrical Engineers in 1964 and became a member of the Council of that institution in 1967. In 1968 he was chairman of Council of The Royal Television Society; in 1955 he had delivered the Fleming Memorial Lecture of that society. The climax of his career was the delivery of the Christmas Lectures of The Royal Institution in 1972. Shortly after this he was awarded the C.B.E., appointed as Visiting Professor of Electrical Engineering at Imperial College of Science & Technology with effect from 1st October 1973, and was elected to membership of The Athenaeum. He was a very gentle man whose enthusiasm and desire for deep understanding of natural philosophy inspired all who came into contact with him.

Letters to the Editor

The Editor does not necessarily endorse opinions expressed by his correspondents

Feedback amplifiers

Following the interesting and informed article by Mr Walker on low noise amplifiers (*Wireless World*, May 1972) there has been a protracted and inconclusive series of letters discussing the various merits of shunt and series feedback connections with regard to noise and distortion.

I would almost certainly have been happy to let this die out in its own way had not the discussion gone completely off the rails in John Linsley Hood's letter "Feedback Amplifiers" in the May 1973 issue.

Mr Linsley Hood suggests that the difference between the series and shunt feedback connections in the circuit given arises because in the series feedback case the signal is not normally attenuated much between source and amplifier, whereas in the shunt feedback case it will be attenuated 4-6dB depending on suitable operating parameters.

The effect of a finite input impedance in a feedback amplifier can be considered as a reduction in loop gain, and for the two connections, see Fig. 1, the effect of input impedance are as below.

Series feedback:

$$\frac{E_0}{E_1} = A(s) \left[\frac{R_e + R_{fb}}{(R_e + R_{fb}) \left(1 + \frac{R_e + R_s}{R_{in}} \right) + A(s)R_e} \right]$$
When $R_e = R_s$ and $R_{in} \to \infty$, $A(s) \gg 1$

When $R_e = R_s$ and $R_{in} \to \infty$, $A(s) \gg 1$ $= \frac{R_e + R_{fb}}{R_e}$

Shunt feedback:

$$\frac{E_0}{E_1} = -A(s) \frac{R_{fb}}{(A(s)+1)R_s + R_{fb}(1+R_s/R_{in})}$$
In the limit $R_{in} \to \infty$, $A(s) \gg 1$

$$= -\frac{R_{fb}}{R_s}$$

It can be seen that the sensitivity of the two circuits to finite input impedance is similar, with suitable values, e.g. a loop gain of $500 R_{in} = 15 \mathrm{k}\Omega$, $R_s = 50 \mathrm{k}\Omega$, $R_{fb} = 500 \mathrm{k}\Omega$, the reduction in gain in each case by considering R_{in} finite is: series 1.3dB, shunt 0.8dB.

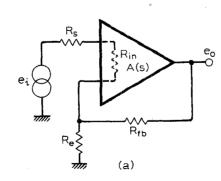
It is not correct to assert that the intrinsic problem with a shunt feedback amplifier is

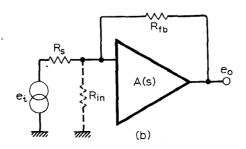
that its input impedance attenuates the signal by 4–6dB. It is readily seen from the equation that the input impedance becomes insignificant anyway when $R_{in} \rightarrow R_s$ and completely insignificant for $A(s)\gg 1$. It is therefore simply a problem of good design to assure that R_{in} is a suitable value, not a drawback of a feedback connection.

Mr Linsley Hood goes on to say that (in the shunt feedback case) the noise impedance seen by the input is not the input resistor circuit value but the value of the "virtual earth impedance", and suggests that this impedance is 600–1200 ohms. This comment is quite amazing. What is a virtual earth impedance? One can only assume that it is a phantom idea to describe how "earthy" the virtual earth point is.

It is quite misguided to use this idea. The virtual earth is a phenomenon resulting from the feedback connection but it does not have an impedance as such that can generate noise.

A shunt feedback amplifier is a current amplifier, and the low noise condition is with the input open circuited, i.e. in Fig. 1(b) the generator E_1 is open-circuited. The noise of the amplifier here is determined by the thermal noise current generator in R_{fb} and the noise factor of the amplifier with a source resistance of R_{fb} .





In its mode of use $(E_1$ short circuited) the source resistance is $R_s//R_{fb}$ and the noise current of R_s is significant. Certainly a $47k\Omega$ source resistor will generate a noise voltage of $3.9\mu V$ and provided the input is short circuited this will be shown in the amplifier noise performance.

In the case of the pickup amplifier $R_s = 47k\Omega$, $Z_{fb}(s) \gg R_s$ it can be shown by calculations that the maximum s/n ratio with a cartridge connected is 58dB ref. 2mV.

Experiment and theory clearly show a marked increase in the noise of such an amplifier when the input is short circuited. Perhaps Mr Linsley Hood could explain how connecting a $47k\Omega$ resistor in parallel with a 1000Ω virtual earth impedance can give a 10dB rise in noise?

Finally, on the subject of distortion, good circuit design can easily permit a series feedback amplifier to have a repeatable performance of s/n better than 70dB s/c ref. 2mV and distortion less than 0.01% in the audio range. The fact that this cannot be achieved with a 741 should be considered irrelevant by any engineer concerned with these and any other important design parameters not covered in the arguments to date. J. R. Stuart,

Lecson Audio Ltd., St Ives, Huntingdon.

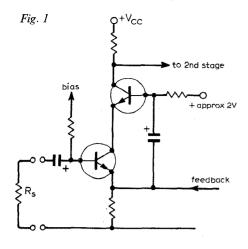
May I offer the following points regarding recent correspondence on distortion and noise?

(1) "Common-mode distortion". In many cases of practical interest, it is the variation of C_{cb} of the first transistor with V_{cb} (Early effect) which dominates. Considering a BC214 input stage run at $V_{ce} = 5$ and handling an input of 1 volt r.m.s., the Texas data sheet indicates a capacitance swing of 3pF. This corresponds to a second harmonic distortion of 0.1% at 20kHz if the source impedance is $10k\Omega$. A considerable reduction in Early effect distortion, and almost complete elimination of the other distortions which are not amenable to reduction by feedback, may be obtained by using a bootstrapped cascode arrangement (Fig. 1).

There is now an Early effect from the upper transistor's C_{cb} , but it is much less than before since it injects distortion into the output, not the input. In fact if the quiescent current through the transistors is chosen for optimum noise figure from R_s , then the Early effect will be reduced by a factor $\sqrt{\beta}$. This circuit permits the lower transistor to be run at a very low V_{ce} , for optimum noise performance, without compromising the ability to handle large signals.

(2) Reduction of distortion by feedback. The statement by Mr Hood and quoted by Messrs Mornington-West and Vereker (May issue), that quadrature components of the feedback are ineffective in reducing the distortion, is absolutely without foundation, as is shown in the appendix to this letter.

To understand the poor high frequency performance of Mr Linsley Hood's *Hi-Fi* News design it is sufficient to consider how much feedback is applied round the output

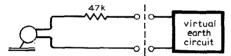


stages. Apart from the usual local feedback, it amounts to $5\frac{1}{2}dB$ at 20kHz.

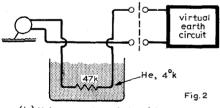
It is not nowadays safe to assume that the effect of a "h.f. stabilising capacitor" is confined to high frequencies. In the design mentioned above, the dominant lag capacitor looks harmless enough at 220pF, yet it gives an open-loop break point of 10Hz.

(3) The measurements reported in Mr Linsley Hood's second letter in the May issue point to interesting possibilities in noise reduction.

Consider a notional dividing line between the $47k\Omega$ resistor and the rest of Mr Linsley Hood's virtual earth circuit (Fig. 2 (a)). The combination on the left will, as he says, produce an open-circuit noise of 3.9μ V.



(a) Quiet record room temperature resistor



(b) Noisy record, cooled resistor

Now let us take a gramophone record on which, by some mischance, tape hiss has been recorded, and let us choose a pickup of the right sensitivity so that this hiss appears as exactly $3.87\mu V$ on its output terminals. Pickup and resistor will then produce $\sqrt{3.9^2 + 3.87^2} = 5.5\mu V$ of noise, but if we now immerse the resistor in a dewar of liquid helium (Fig. 2 (b)) the open circuit noise will once again be $3.9\mu V$.

The impedance presented to the circuit on the right is of course exactly the same for Fig. 2 (b) as for 2 (a), so it should need only a little fiddling of the frequency spectra to convince the circuit that it is connected as in Fig. 2 (a), when the truth is 2 (b). If now the circuit can achieve the noise value of $0.6\mu V$ claimed by Mr Linsley Hood, then a noise reduction of $20 \log_{10} \frac{3.87}{0.6} = 16.2 dB$

will have been obtained.

Perhaps some enterprising record company will consider this technique for revitalising its pre-Dolby LPs?

Peter G. Craven,

Oxford.

Appendix

Let the amplifier have perfect common mode rejection so that V_{out} is a function of V_1 only. Suppose that we are trying to reproduce a sine wave of unit amplitude and that it is possible to predistort V_1 so that V_{out} is a pure sinusoid. Let "X" be the assumption that the gain of the system to a small signal superimposed on the input is not greatly affected by the presence of the large signal. X will be false if the amplifier is near to clipping.

Let the gain (V_{out}/V_1) of the amplifier at the *i*th harmonic of the sinusoid be A_i and let β_i be the corresponding feedback factor (V_f/V_{out}) . To take account of phase shifts, A_i and β_i will be complex. Suppose that we have succeeded in making V_{out} a pure sinewave and that d_i (also complex) is the amplitude of the *i*th harmonic of the predistorted signal V_1 necessary to achieve this. Since the feedback network is assumed linear, V_f will be a pure sine wave, and since $V_{in} = V_1 + V_f$, it follows that V_{in} must also have an *i*th harmonic of amplitude d_i .

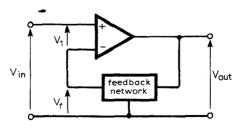
We wish to consider a pure V_{in} , and this we get from the predistorted V_{in} by adding $-d_i$ of the *i*th harmonic for $i=1\ldots\infty$. By assumption X there will appear harmonics at the output, the amplitude of the *i*th being $-G_id_i$, where G_i is the gain of the system (including feedback if any) at the *i*th frequency.

Since d_i does not depend on the feedback, we have proved the well known fact that feedback reduces each distortion product in the same ratio as it reduces the gain at the frequency of the distortion product.

By elementary feedback theory, G_i is given by

$$G_i = \frac{A_i}{1 + A_i \beta_i},$$

and comparing with the case $\beta = 0$, it is clear that introducing the feedback has reduced the gain, and hence the distortion



by a factor $|1+A_i\beta_i|$. This factor we now evaluate for an amplifier with a loop gain of 10 ($|A_i\beta_i| = 10$).

No phase shift $\rightarrow A_i \beta_i$ is real and positive $\rightarrow |1 + A_i \beta_i| = 11$ 90° phase shift $\rightarrow A_i \beta_i$ is pure imaginary.

90° phase shift $\rightarrow A_i \beta_i$ is pure imaginary

 $\rightarrow |1 + A_i \beta_i| = 10.05$ 180° phase shift $\rightarrow A_i \beta_i$ is real and negative

 $\rightarrow |1 + A_i \beta_i| = 9$

Mr Linsley Hood replies:

I am sorry that Mr Stuart feels that the debate on feedback amplifiers "has gone completely off the rails", but he has taken my letter somewhat out of its intended context.

To refer specifically to the main point of this—measurements suggest that the s/n ratio of an amplifying circuit with shunt f.b. is a few dB worse than in the case of the series f.b. circuit with the same value of input resistance.

I believe that this phenomenon is real, and that it is due to the fact that any real amplifying device will require some input energy—significant in a bipolar transistor—and that in the shunt circuit this is obtained from the input signal.

In the latter part of my letter I suggested an alternative method of considering the noise impedance seen at the input—which is a voltage node—in a shunt f.b. amplifier. If one considers the amplifying element, having a known open loop gain, as being detached from the feedback loop but amplifying the noise voltage seen at that point, the noise impedance of the "virtual earth" can be derived, if one is interested to do this.

This observation was not specifically related to the s/n ratio of a shunt feedback circuit, which is best approached by considering it as a current amplifier. In this case the input noise currents decrease as the root of the admittance (1/Z) of the input limb, whereas the signal current decreases linearly. Other things being equal the lower the input limb impedance, the better.

In the particular case of a pickup amplifier circuit with R.I.A.A. equalisation, it should be remembered that the effective noise bandwidth is only about 500Hz. Since this allows a s/n ratio with a $47k\Omega$ input resistor and a shunt f.b. circuit to be -72dBref. 5mV (-64dB ref. 2mV) I suspect that the "calculations" to which Mr Stuart refers assume a wider bandwidth than this. The relative advantage of the series circuit diminishes with frequency when used with an inductive element such as a magnetic p.u. cartridge, from about 11dB at 1kHz to some 3dB at 5kHz. (Assuming a 600mH cartridge inductance, and a series f.b. input d.c. resistance of $2k\Omega$).

In reply to Mr Craven, on the more important point of the extent of distortion reduction by feedback at phase angles other than 180°, the problem is that the predicted distortion reduction from the formula

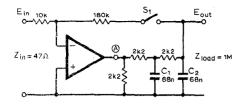
$$\frac{D_F}{D} = \frac{A_F}{A}$$

gives unsound results under these conditions, whether the gain is calculated by the method Mr Craven shows or whether it is derived by the classical formula below.

$$A_F = \frac{A}{\sqrt{1 + |\beta A|^2 - 2|\beta A|\cos\Phi}}$$

where Φ is the f.b. phase angle.

As an example, a non-linear amplifier element, having a gain of $100 \times$, an input impedance of $4.7 k\Omega$ and a t.h.d. of approximately 4% at 1kHz and 1.5 volts r.m.s. output, was set up as shown in the figure, with an output lag circuit whose values were chosen to give a phase lag of 90° at 1kHz.



When S_1 was closed, to apply "negative" feedback (10dB at 20Hz, although the gain reduction was only 2dB at 1kHz) the t.h.d. at 1kHz increased from 4.7% to 6.3%. However, when the phase shift introducing elements (C_1 and C_2) were removed, closure of the switch to introduce true negative feedback reduced the t.h.d. from 4.1% to 1.3%, which is in line with theory. In both cases, when feedback was applied, the input to the amplifier was increased to give the same output at point "A" (1.5V r.m.s.) as in the no-feedback case.

J. L. Linsley Hood, Taunton, Somerset.

Car seat belts

As a campaigning member of the relatively small percentage of seat-belt wearing drivers, I nevertheless must add a note of protest against the gleeful hand-rubbing in anticipation of forthcoming contracts which formed the basis of Vector's article in the May issue on the proposed automatic seat belt interlocks involving the use of i.cs.

The i.c. boys have indeed "got it made" as he suggests, and advantage will doubtless be taken of the opportunity to scale new heights of technological overkill with the machines on which many are obliged to depend for personal transport. One needs only to observe the panic of the visitor to London when confronted with automatic ticket barriers at tube stations to judge the total inadequacy of the non-technical majority when faced with mechanisms which demand a specific sequence of human responses in order to perform their function.

I realise that humans adapt gradually to the numerous complexities of modern living, but have you ever tried to fasten an American 3-strap seat belt while lights flash and a mind-numbing buzzer sounds? The experience is the most persuasive inducement to leaving the belts permanently locked behind the seats.

Quite apart from this, many people firmly believe that the individual should retain the right to be hurled through the windscreen of his car at will, and that the burden of his personal safety in this instance should neither be shifted to the already over-intrusive state, nor to the already overworked police force.

Any safety regulations and technological advancements which prevent an individual from endangering others must be applauded, but regulations covering personal safety could never be far-reaching enough to be 100% effective, and until such time as you produce a solid-state system as a substitute for the

human brain — there will always be a man prodding wires into an electrical socket with matchsticks, leaping from a bus into the traffic stream, and refusing to wear his seat belt.

Lyn Heigl, Studley. Wilts.

Magnetic units

I am much obliged to Dr McCaig for replying in the June issue so informatively, direct from the Permanent Magnet Association at Sheffield, to my articles on magnetism (Jan. and Feb. issues). (At least the P.M.A. hasn't made complete nonsense of my criticism directed at Sheffield by having moved, unknown to me, to Bognor Regis or Lerwick, in keeping with the trend towards dispersal.) My apology is due to the P.M.A. for my having suggested that it, like the incredible referee cited by Dr McCaig, is still unaware of SI units. I do, however, remind readers that my technique of trailing my coat provocatively in order to elicit information is or should be well known by now. And I wish Dr McCaig had made clear what practice has been continued in the majority of P.M.A. publications during the last 20 years. Listing permanent magnet properties in m.k.s. as well as c.g.s. units? If so, it would seem to be as if our Government were to continue for the 20 years after 1971 to give money values in decimal units as well as £.s.d. In short, not much encouragement to make the effort to change over. Which may be one reason why the majority of papers on magnetism continue to use c.g.s. units. (This was actually my essential grouse.) And what units do individuals within the P.M.A. normally use?

However, assuming that the P.M.A. collectively and individually is doing all it reasonably can to encourage abandonment of c.g.s., with renewed apologies to it, I hereby amend my strictures by altering the address of the target from Sheffield to all those places where people use c.g.s. units without reasonable excuse. In this at least I would seem to have the influential backing of Dr McCaig.

Although I'm not altogether convinced that standardizing the unit and symbol for "intensity of magnetization" (or whatever) would make all the difference to acceptance of SI units by workers in magnetism (surely the question arises with any system of units?), I'm at one with Dr McCaig on the desirability of tidying up here. In so far as I have used this quantity (and that is not at all far, as I shall explain in a moment) I have used the symbol M, but I would happily accept J if that is decreed; in fact, I'll accept it now, for although there is unlikely to be any confusion with mutual inductance they do both come into related contexts. And my preferred unit is the tesla, since from a practical point of view one can regard J as the difference between the actual flux density and the corresponding flux density in vacuo, both of these being in teslas. But I take Dr McCaig's point that if the chosen measurement technique makes it more appropriate to treat it as a kind of H by reckoning it in A/m, fair enough.

Finally, I must reply to the charge that I ignored the need for the quantity J. I plead guilty. But without remorse. For Dr McCaig is a physicist and represents magnet manufacturers. I am an engineer and, for the purpose under discussion, purport to represent consumers of magnets among a great many other things. So it is hardly surprising if the attitudes to J are quite different. I don't for a moment doubt that to the people Dr McCaig represents J is, as he says, fundamental. Under a disguise, for readers who I assumed were interested in the theoretical physics of magnetism, I myself once found it necessary to introduce J (as M). But I very much doubt that it is a necessary or desirable part of an elementary 7-page treatise on magnetism and magnets for readers to whom this subject is incidental. I've never once needed J to deal with practical problems roughly within the field covered by Wireless World - in which of course the word "Wireless" is rather like the "candles" allegedly brought into the House of Commons when it is too dark for Hon. Members to see their order papers clearly. It is interesting to note that although the physicists Bleaney and Bleaney in their substantial volume Electricity and Magnetism (using m.k.s. units) deal with J (as M, which is probably where I got it) in the standard theory, they hardly mention it in the sections on magnetic measurements. And in Hvistendahl's book on units, J (or anything like it) and susceptibility are not mentioned. But then he is an engineer. "Cathode Ray."

Printed circuits the easy way

Most people who have attempted printed circuit work of a one-off nature will admit this can be time consuming and tedious using the normal method of draughtsman's pen together with either brushing Belco or Humbrol as a resist. It requires considerable artistic acumen or the patience of a saint!

This chore finished and drying time allowed, one has then to wait somewhere between 25 and 50 minutes while surplus copper very slowly dissolves in the etching fluid.

The result of all this time and trouble ought to be a neat web of sharp unbroken lines and sensibly circular drilling points, not the sorry bodge which frequently causes the experimenter to revert to tag board or pin board.

There had to be an easier way and out of sheer frustration the following was evolved. Instead of the draughtsman's pen and Belco a fibre tipped waterproof marker pen costing about 22p, called the 'Miracle Pen' (of Japanese origin) was adopted and is a delight to use, producing fine or thick lines with ease. (A 1in \times $1\frac{1}{2}$ in board took 10 minutes to draw,

drill and etch.) Virtually no drying time is required. The pen is refillable, but should last a considerable time provided the cap is replaced when not in use. For professional looking numbering and lettering the self adhesive 'Letraset' used as a resist has proved most effective.

All that remained was to speed up the etching. Apparently some form of effective agitation was required. Eccentric cams were suggested for wobbling the tank but just appeared to cause swirl; supersonic agitation was obviously out for kitchen use. The answer proved to be aeration; this was easily and cheaply provided by an electric aquarium air-pump costing £1 together with a porous air-stone at 5p to disperse the bubbles. A glass water jug, narrow at the bottom and wide at the top, was half filled with a strong solution 60/40 ferric chloride plus 2cc hydrochloric acid*. This is diluted with an equal quantity of water. The jug was placed in a polythene washing-up bowl full of fairly hot water in order to warm the solution the air-stone was placed at the bottom of the jug and connected by its tubing to the aerator.

When switched on, due to the "Vee" shape of the tank, even dispersion of bubbles throughout the fluid was observed. The work to be etched was then suspended in the fluid, etching was completed in five minutes.

Rinse under the cold water tap to remove etching fluid, then remove resist with rag well moistened with cellulose thinners. A final rub with Brasso and the process is complete.

Should you make a mistake when drawing the circuit a rag moistened with cellulose thinners can be used as an eraser.

One other point I feel worth mentioning. After assembly and soldering it is worth while rubbing over with cellulose thinners to remove untidy flux, then applying a coat of Ronseal thinned with slow cellulose thinners as protection.

J. Ferguson, Penrith. Cumberland.

V.H.F. receiver performance

Isn't it time that some "figure of goodness" was instituted for v.h.f. receivers, other than the useless "locks in at $\frac{1}{2}\mu$ V" or whatever?

It is a revealing experience to try several different v.h.f. sets on the same aerial. The results - especially on stereo go all the way from very good to ruddy 'orrible! Yet the specs give no hint at all as to which is which. Nor does a list of semiconductors give any indication.

I own two v.h.f. stereo receivers, one all-germanium with a stereo lock in of 5 μ V, lineup r.f.-osc/mixer-3 i.f.-ratio disc; the other all-silicon with a stereo lock in of $\frac{1}{2}$ μ V, lineup f.e.t. r.f. and mixsep-o.s.c-2 i.c. (CA3012)-Foster-Seely disc.

Now which would you expect to be the best? The actual fact is the former! (Though I'm working on the latter.)

Point being that in an advertisement it would be the i.c. one which would seem to be the best.

Given a dozen units to choose from how do you do it? Advice to "try" is absurd. Surely there should be some British Standard figure that could decide. Ronald G. Young,

Peacehaven.

Sussex.

Amateur computer club

I feel that you and your readers may be interested to hear that, as a result of advertisement in your journal and other magazines last year, the Amateur Computer Club has been formed.

The club is for those interested in the construction, design or programming of computers as a hobby. At present the main activity is the production of a newsletter (which appears every two or three months) which acts as a databus to distribute information on hardware and software techniques of interest to the members.

Anyone interested in the club may receive further details from me at the address below (s.a.e. appreciated).

M. Lord. 7 Dordells, Basildon, Essex.

Power supply design

I was interested to read Mr R. Aston's article in the May issue. The idea of a switching pre-regulator controlling the voltage across the series element in the main control loop is not new.

Previous designs in my experience have mostly used thyristors operating at line frequency for the pre-regulator. I was myself involved in such a design for an ultra stable (5 p.p.m.) wide range constant current supply. For reasons of isolation this was operated from a 400Hz motor generator set, but our experience showed that it would operate equally well at 50Hz with suitable adjustment of smoothing components.

The arrangement shown below in the

diagram was used. A similar system is to be found in power supplies manufactured by the Harrison Division of H.P. and elsewhere.

With the advent of modern power switching transistors an obvious modification to the above circuit would be their use in a switching pre-regulator. However, I was much discouraged in this line of development by the manufacturers (including Mr Aston's company) of commercial switching supplies who were not interested in my proposal for a system almost identical to Mr Aston's.

J. F. Hilev. Bishop's Stortford, Herts.

Audio amplifier design

I read with interest the letters on audio feedback-amplifiers published in the May 1973 issue (Letters pp. 246-248), and would like to comment on one or two of the points raised.

Messrs Mornington-West and Vereker make a useful point regarding the noise contributions due to later amplification stages. While I agree with their analysis, it may be difficult to apply the result in the case when stages are current rather than voltage driven.

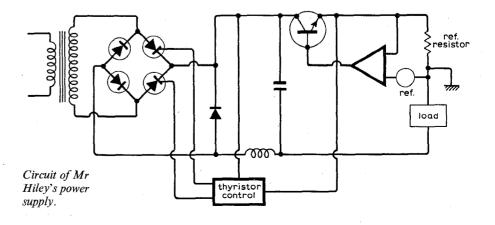
This is the case for the second stage of the amplifier shown on p. 236 of my article on low-noise amplifiers (May 1972); I found that current driving gave better linearity in large-signal stages. For these conditions, emitter degeneration in the second stage is undesirable, and in any case major-loop feedback is preferable provided stability margins are met.

The noise contribution of the second stage is readily calculated using Faulkner's concept¹ of noise resistances, and since the second transistor is being current driven only the noise current generator need be considered. Its noise is transformed via the input transistor (a voltage to current converter) to a noise voltage source or series noise resistance at the input, of value:

$$R_{nv2} = g_{m2} \cdot \frac{(1 + g_{m1} \cdot R_{e1})^2}{2 \cdot h_{FE2} \cdot g_{m1}^2}$$

Reference

1. Faulkner E. A. "The Design of Low-noise Audio-frequency Amplifiers", The Radio and Electronic Engineer", vol. 36, 1968, pp. 17-30.



^{*}The chemicals used by the writer are dangerous and should be treated accordingly. Ed.

Comparing this with the series noise resistance due to collector current shot noise in the input transistor $(R_{nv1} = 1/2 \cdot g_{m1})$, we see that for R_{nv2} to be negligible (say one tenth),

$$g_{m2} = \frac{h_{\text{FE2}} \cdot g_{m1}}{10 \cdot (1 + R_{e1} \cdot g_{m1})^2}$$

hence a ratio of collector currents of

$$\frac{I_{C2}}{I_{C1}} = \frac{h_{FE2}}{10 \cdot (1 + R_{e1} \cdot g_{m1})^2}$$

This is step 4 in the design procedure given in my article (p. 236).

In practice, the second transistor will contribute excess noise in its noise current generator so a good margin of safety is advisable in shot noise calculations. For the feedback triple given as the example, the second stage contributes a noise resistance of about 60 ohms which is negligible compared with the series resistors and noise resistance in the input circuit. The presence of flicker noise in the second stage may be detected by shorting the input of the amplifier, thus nullifying the effect of excess noise in the first-stage current generator.

I do not wish to reiterate the contents of my article or previous correspondence in connection with Mr Linsley Hood's reply to my April letter, but from a simple "ideal case" analysis it is clear that with a lowimpedance source (e.g. a pickup cartridge) the parallel termination offers a very much better noise figure than the series resistor. This follows directly from the noise figure

N.F. =
$$10 \log_{10} \left(1 + \frac{R_{series}}{R_{source}} + \frac{R_{source}}{R_{parallel}} \right) dB$$

and is independent of any attenuation due to finite input impedance of the amplifying device (intrinsic or otherwise).

In fact the attenuation of 4-6dB, quoted by Mr Linsley Hood, due to this cause, is quite incompatible with the achievement of an adequate noise figure which demands a common-emitter transistor input impedance several times greater than the source impedance, i.e. voltage drive conditions (Ref. 1, section 3.2). If he believes the input attenuation to be the problem, why does he not use a field-effect transistor?

When he refers to the "noise impedance" seen by the input does he mean the equivalent noise resistance? Consider a simple inverting operational amplifier of gain A with equal feedback and input resistors and with the input grounded. By summation of currents at the virtual earth, the noise output voltage is $\sqrt{4kT\Delta f\cdot 2\bar{R}}$ giving a voltage at the amplifier input of $(4kT\Delta f \cdot 2R)/A^2$ mean square volts, or an equivalent resistance of $2R/A^2$. This disagrees with the virtual earth impedance of approximately R/A, the summing point in fact being quieter.

One would not expect the noise at the virtual earth to correspond to the noise voltage across the input resistor since the amplifier is sensing the currents flowing. The presence of a virtual earth (since it is due to n.f.b.) does not affect the noise figure of the input transistor which should be optimised under open-loop conditions

(not for the virtual earth impedance) and the feedback used to alter the input impedance (Ref. 1, section 4.1). The virtual earth impedance does not enter into the noise calculations since it is a function of the open-loop amplifier gain which can be chosen arbitrarily from noise considerations.

H. P. Walker, South Queensferry, West Lothian.

I thank Mr Walker for his further letter, and concur in general with his comments.

I agree that a high input impedance device such as an f.e.t. would avoid the problem of input energy loss on voltage to current conversion. The difficulty is that the required input impedance for such a device, for optimum device noise figure, is in the megohm range, which is unsuitable for a low input impedance system.

J. L. Linsley Hood,

Taunton.

Somerset.

Distortion reducer

Mr Bollen, in the April issue (Letters, p. 192), has correctly pointed out a term which I omitted in my analysis of his Distortion Reducer. Nevertheless I should like to develop the idea that his system is essentially equivalent to the procedure of increasing conventional negative feedback around the main amplifier and adding a pre-amplifier.

The correct expression for the input to the main amplifier in the basic Distortion Reducer is -S-S+S+D. With a gain G_2 in the distortion channel this becomes $-S-G_2S+G_2S+G_2D$, which we may write as $-S(1+G_2)+G_2(S+D)$. The first term is the original signal multiplied by the gain of the equivalent pre-amplifier, and the second term is equivalent to that produced by a conventional negative feedback path.

I also offer the following more general treatment, in the hope that it may be productive and begin to answer Mr Cocking's call for a comparison of the two ways of reducing distortion.

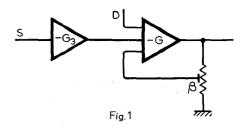
Let the main amplifier have a (complex) gain -G, the term D now being reserved for distortion other than a mere quadrature term, let the attenuation at the output (previously 1/G) be A, and let the distortion channel have a (real) gain G_2 , as before. Writing x for the net input to the main amplifier, following the signal round the loop, and solving for x, we find

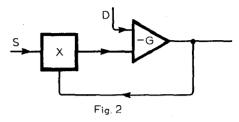
$$x = -\frac{S(1+G_2)+D}{1+G_2AG} \tag{1}$$

If G is real (no phase shift in the main amplifier) and A is made equal to 1/G as before, this reduces to

$$x = -S - \frac{D}{1 + G_2}$$

confirming the earlier result that the equivalent distortion input D is effectively reduced by a factor $(1+G_2)$.





Now consider the conventional circuit shown in Fig. 1. A similar analysis leads to

$$x = -\frac{G_3 S + D}{1 + G\beta}$$

If β is chosen to be equal to G_2A , and G_3 to be equal to $(1+G_2)$, this becomes identical to eqn. (1), showing that the two feedback systems are equivalent, even when there is a phase shift in the main amplifier, and that stability should be as good, or as bad, with either system. The relative merits of the two systems may turn out to depend on subtle considerations of linearity in the op-amps, interference with the input circuit of the main amplifier, or the like.

Finally, I offer Fig. 2 for consideration. X is a linear active network. It will be seen that the Distortion Reducer, the conventional circuit, and others as yet unconsidered can all be drawn in this form. One would expect them all to behave similarly. Richard G. Mellish,

Watford.

Herts.

Quantity names

Since the discussion on my "Unified Dimensional Display" (March 1972), was published in the January 1973 issue, the dimensional position occupied by "magnetic vector potential" has been allotted the name "fluxivity" by John G. McKnight in a project note on test tapes appearing in the Journal of the Audio Engineering Society for March 1973. This is in order to describe the quantity "tape flux per unit track width" used in magnetic recording, measured basically in Wb/m, but conveniently referred to nWb/m in actual practice.

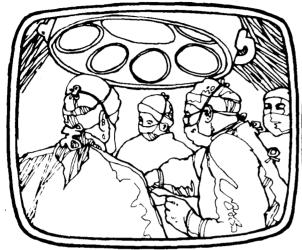
In view of this tendency to evolve new quantity names when needed, I suggest the replacement of "1/permittivity", or "reciprocal permittivity", (unit m/F), in my display system by the quantity "forbidivity" (or "forbiddivity"), forbid being an antonym of permit - or can readers suggest something better?

R. N. Baldock,

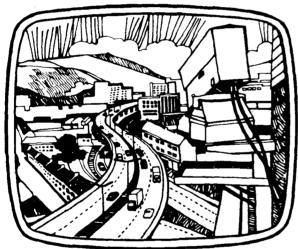
Harrow.

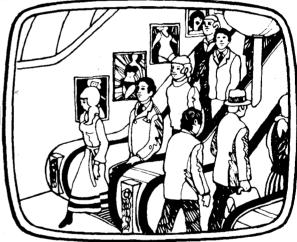
Middlesex.

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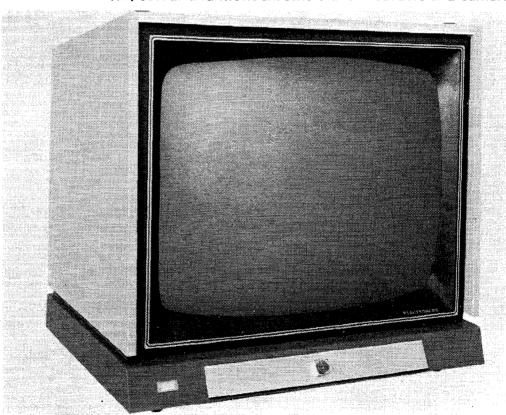
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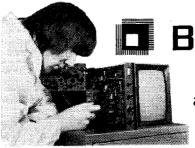
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Montreux Television Symposium

Impressions of the conference, papers and exhibits

by Harold Barnard

It is twelve years since the first International Television Symposium was held in Montreux — originally as part of the Golden Rose television programme festival. There was a very small associated exhibition — little more than baize-top table displays — and 200 or so delegates attended.

What has made this biennial event the Mecca for so many of the world's manufacturers and users of television equipment? It is certainly not the quality of the 100 or more papers presented during this year's meeting (18-24 May). There were undoubtedly highlights, but many of them were little more than sales blurbs on equipment on show. The associated (or perhaps one should say "dominating") exhibition is now indisputably the largest in Europe devoted exclusively to television equipment. Then, of course, there is the venue. How big a part this plays in influencing nearly 2,000 delegates to register (several hundred of them accompanied by their wives), one can only hazard a guess. In all, of course, the delegates and their wives, plus the 500 or so staff of the exhibitors, must have brought a considerable income to the municipality of Montreux who, in collaboration with the Swiss F.T.T., are the organizers. Perhaps here lies the secret the event is publicized by and has the financial backing of the town's tourist

Some measure of the importance attached to the occasion can be judged from the fact that the delegates came from 30 or more countries and that there were special delegations from Australia (over a dozen engineers) and South Africa (nine); each of which will be starting colour services in the next few years.

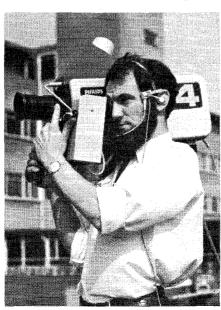
The exhibition occupied four floors of a recently completed conference and exhibition centre. There were nearly 100 exhibitors from 15 countries; among them 22 from the U.K., 18 from Germany, 16 from Switzerland, 13 from France and 11 from the U.S.A. It should, however, be added that several of the companies shown in the catalogue with addresses in Europe, are in fact, subsidiaries of American concerns. It was gratifying to see so many of the smaller specialist U.K. companies, such as Link Electronics, Crow of Reading, Engineering Designs & Supplies, Prowest Electronics, System

Video, and Michael Cox Electronics, competing with the big battalions. Incidentally, when reporting on the second symposium (1962), we recorded that there was only one piece of British Equipment on show and that was part of an American unit!

From what has been said it will be blatantly apparent that the raison d'etre was the exhibition. The show and the symposium are complementary but if the organizers are desirous of maintaining the position the occasion has gained, they will have to look more closely at the quality of papers accepted. The criteria should be fewer and better papers. The standard of the papers presented at the biennial London International Broadcasting Convention is much higher, but the I.B.C. exhibition has not so far acquired the international prestige accorded to the Montreux event.

Before dealing with some of the new equipment shown, we will look at a few of the papers presented.

The two sessions (there were 13 in all) which drew the largest attendance of delegates were those devoted to cable television. At one of these, specialists from various countries and international bodies



Prototype of the Philips LDK15 portable colour camera, demonstrated at Montreux, which is fitted with a combined 1-in Plumbicon tube and image intensifier.

presented reviews of the present position of cable television, or CATV (common antenna television) as it is more commonly called both in Europe and the U.S.A. These were preceded by a survey of the legal problems encountered in the distribution of programme material over cable networks; for example copyright, artists' fees. The speaker, G. Straschnov of the E.B.U., suggested that "the law is always far behind technology".

Another contentious aspect of cable or relay distribution services is that of programme origination by the operators. If, as in the U.K., operators are hamstrung by regulations limiting the material to be distributed, what chance is there of CATV becoming the multi-channel information/entertainment service envisaged by W. J. Bresnan, of the National Cable Television Association, Washington? He foresaw CATV as a two-way public access system providing not only entertainment and general information, but an educational and advice service. On the question of the number of channels envisaged it was pointed out that in California there is a 40-channel system operating and 11 channels are in use in parts of Belgium where programmes in three languages from neighbouring countries are relayed. According to figures, quoted by Bresnan, seven million of America's 65 million television households receive their programmes via cable.

Several of the formal papers presented at the CATV session were concerned with the need for a two-way data communications capability, linking each subscriber with the system terminal station and /or other service centres, e.g. schools, banks, stores, information centres, etc. As one author expressed it, CATV is no longer a means of providing television reception for viewers in areas of poor or indifferent signal levels, but has come "to the point where we now stand at the threshold of the wired city and the wired world". This was said by Selig Lenfesky, of Coral Inc., whose paper was concerned with bi-directional amplifiers for CATV. The scheme he outlined provides for 31 television channels, each of 6MHz bandwidth, plus a 20MHz band for f.m. sound channels and two pilot carriers in the "forward" direction and, in the return direction, two 6-MHz vision channels and many data channels.

Several of the papers on CATV dealt with specific distribution problems. For instance H. G. Schwarz (RCA) presented a very detailed quantitative analysis of the selectivity requirements necessary in a receiver to avoid visible interference from signals propagated on adjacent channels. Whereas in television broadcasting adjacent channel interference can be alleviated, if not cured, by careful geographic distribution of frequencies, in cable distribution systems signal levels of adjacent channels are nominally equal. It was, in fact, found that the requirements for vestigial sideband attenuation for cable system modulators is considerably more stringent than those established by the F.C.C. for American broadcast transmitters!

Another RCA Laboratories paper (presented by C. M. Wine) described a method of two-way data communication for ancillary services using a cable television system. It is based on polling individual subscriber terminals. Downstream messages consist of a string of bytes or words sent asynchronously at a data rate of 1.25 Mb/s. Each byte consists of a start bit, seven data bits, a parity bit and a stop bit, and has a duration of 0.8µs. The downstream message has the following format: SYN, TID1 TID2 DID, and arbitrary number of data bytes, SYN. The initial and terminal SYN characters indicate to the terminal the start and end of the message. TID1 and TID2 identify the particular terminal being addressed; this allows 16,384 terminals on the system. DID identifies which one of 127 possible service-related devices is being specified, e.g., the security system, an opinion poll keyboard, a teletype terminal, a credit card reader, etc. The data bytes, if required, are then routed to that device.

Upstream messages also consist of a string of bytes sent asynchronously at 1.25 Mb/s. The upstream transmission commences immediately after TID2 (as soon as the particular terminal has been identified). An experimental laboratory version of the system operates on a test-bed v.h.f. cable television system. Data is sent using f.s.k. modems at 112-114MHz downstream and 12.3-13.8MHz upstream. The modem data interface is a conventional bit serial logic level interface and is independent of carrier frequency and modulation method. Thus the same data communications approach could be used in virtually any cable environment including an h.f. switched system.

Recent developments in camera tubes were discussed by W. E. Turk (EEV). He dealt briefly with the light bias leddicon tube, which operates on the principle of improving the beam acceptance curve of the target by artificially raising its general potential level. This is done by including within the tube a small light source to produce uniform background illumination. Mr Turk dealt at greater length with the need for more sensitive tubes in monochrome cameras for industrial television He then did some crystal gazing and demonstrated pictures taken with a pyroelectric vidicon. This experimental tube

uses as its sensing target a crystal of triglycine sulphate (t.g.s.) fashioned into a micro-thin slice for mounting as a conventional vidicon target. The degree of polarization of t.g.s. is temperature dependent so that when a thermal image is projected on to it a corresponding voltage pattern is produced on the scanned surface.

There is a constant endeavour to accommodate more and more information in the 5 or 6MHz channels allocated to television. Several examples were discussed or demonstrated. Dr Jorg Irmer, of SEL, Berlin, described a two-carrier technique developed by the Institut fur Rundfunktechnik, Hamburg, which is to be used at the Regensburg u.h.f. station later this year. The normal sound carrier for television systems B & G used in Germany is 5.5MHz above the vision carrier and the output is 10dB below the nominal vision channel power.

The additional sound carrier is located 242kHz above the normal sound carrier and the power level will be 16dB below the vision power. Both sound channels are frequency modulated. In the exciter for the transmitter the audio and f.m. sections are duplicated; the first modulator producing the normal i.f. signal (33.4MHz) and the additional one an i.f. of about 33.15MHz. Both i.f. signals are superimposed in a combiner circuit and then passed to the r.f. mixer/linear amplifier and applied to the klystron power amplifier.

The last, but certainly not the least important, paper to be presented at Montreux was by D. R. Wells, of the Public Broadcasting Service of America. It dealt with a method of transmitting coded caption information (during the vertical blanking period of regular broadcast television signals) for the benefit of viewers with impaired hearing. An adapter at the home receiver decodes the caption information and displays it on the television picture. The system, to be used experimentally later this year by the Public Broadcasting Service in six cities, was developed by the Time and Frequency Division of the U.S. National Bureau of Standards. The N.B.S. has been investigating for some time the possibility of using the American television networks for transmitting precise time and frequency information during the vertical blanking period and, in fact, developed what was called the TvTime system and it was from this that the proposed captions for the deaf emerged.

The proposed signal is transmitted on line 21 of the vertical blanking interval. On odd fields the signal consists of a 1MHz sine wave followed by a 26-bit data stream containing the caption information. On even fields only the 1MHz sine wave is transmitted. It will be possible to transmit about 600 words of captioning per minute. To insure legibility against any background, each letter in the captions, which will be in two rows of capital letters (26 characters per row), has black edges. Even when a picture is noisy, as might be the case in fringe reception

areas, the captions are clear because they are generated by the adapter fitted to the standard receiver. Tests using the N.B.S. signal on line 21 have been no more noticeable to viewers using normal receivers than vertical interval test signals (v.i.t.s.) or the vertical interval reference signal (v.i.r.s.). Apparently most home receivers are said to be adjusted by the manufacturers to be overscanned. Annoyance is only likely if a set is slightly underscanned and the caption information would then appear as a line of moving dots.

So much for the papers. What of the exhibition? It is impracticable to do justice in a short article to the galaxy of new equipment shown — monitors, test and measuring equipment, cameras, control consoles, video and sound recorders, telecine and transmitting gear etc.

Among new equipment shown by some of the smaller British companies mentioned earlier was a range of waveform measuring instruments, including a vector monitor, from System Video Ltd, of Chertsey, Surrey, a new company making its debut at Montreux. Also shown were a range of monochrome monitors and a 17-in colour monitor (Prowest, of White Waltham, Berks); a 14-input vision mixer for mobile use (Michael Cox Electronics, Twickenham, Middx); and a new three-tube colour camera with built-in zoom lens (Link Electronics, Andover, Hants).

Thomson-CSF, of France, featured a new video tape recorder, the TTV 3500, which employs a recording angle of 19°. Developed jointly by IVC in America, Thomson-CSF, and Rank-Cintel in the U.K., the machine uses 2-in tape and has five longitudinal tracks in addition to the video record. The two main audio tracks are separated by what is called an "arrangement track" to eliminate crosstalk. The fourth and fifth tracks serve for drive control and for the encoded address used for achieving accurate electronic settings. The tape is wound at 180° on an air lubricated drum.

With the extension of television broadcasting hours and the use of unmanned transmitters the need has arisen for testing to be carried out during transmission of the programme signal without affecting the viewer's picture. This has led to a change in emphasis from full-field testing to insertion test signal measurements. Marconi Instruments were showing an automatic monitoring system based on these insertion signals. The two instruments forming the new monitoring system are the insertion signal analyser TF 2914, which permits simultaneous measurement of a variety of signal parameters using standard national or international test waveforms, and the data monitor TF 2915 which provides control for the TF 2914 and the facility for comparison of the measured parameter with a pre-determined limit. As a result of this comparison executive action is automatically taken in the equipment as required either to change the transmitter from main to reserve or to select an alternative video input to the transmitter.

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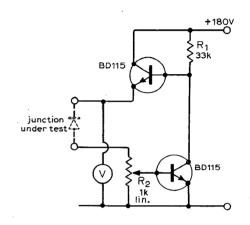
Simple breakdown voltage meter

The circuit will test the V_{CBO} , V_{CEO} and V_{EBO} breakdown voltage of most types of transistors, small signal or power; V_Z of zener diodes and the reverse breakdown voltage of small power diodes. It should fit easily into most instruments which lack this facility with little modification.

Power is supplied by two small 90-V batteries in series. Resistor R_1 biases the upper transistor into conduction and voltage is applied to the junction under test.

As this breaks down current flows in R_2 causing the lower transistor to conduct, with a consequent drop in base voltage of the upper one. Thus R_2 may be used to set I_{BR} over a fairly wide range, enabling the user to study the effects of different current (on V_{CEO} breakdown in particular). The voltmeter reads V_{BR} plus the drop across R_2 which is small and can be neglected. Transistors of type BD115 are used but other high voltage types, e.g. MJE340 could be used.

J. W. Brown, Gateshead, Co. Durham.



Phase-locked loop demodulator

In the November 1972 issue of Wireless World Pat Hawker showed a circuit for a low-cost phase-locked loop demodulator using three digital integrated circuits and two transistors ("Synchronous Detection in Radio Reception-2", page 527). Here is an even simpler circuit which can be built for about a third of the price.

Gates A, B and C constitute a relaxation type of voltage-controlled oscillator whose output frequency is determined by the value of C_1 and positive current sources supplying pins 10 and 13. It should be noted, however, that the inputs to gates A and C supply part and whole of these currents respectively. Similarly, no resistor is required on the output of gate C, pull-up being provided by the input current to gate B.

D is arranged such that when pin 6 is high the gate is biased by R_2 and R_3 to operate as a linear amplifier for the input signal. In operation, however, pin 6 is alternately high and low due to the oscillator output and hence gate D performs as an amplifying phase detector. The output from this stage is fed via the low pass filter R_4 and C_2 to the voltage-controlled oscillator, completing the phase-lock loop. A separate filter R_5 and C_3 provides the audio output.

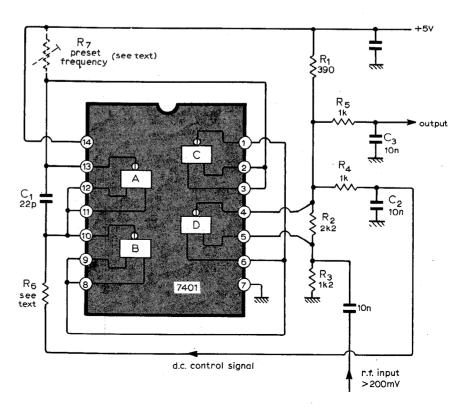
With \hat{C}_1 is equal to 22 pF the circuit operates at a frequency of approximately 10 MHz. Making R_6 270 Ω or $10 \text{k} \Omega$ maintains lock over a frequency range of 2

MHz or 300 kHz respectively. In both cases the output swing is just over 1 volt. An additional component R_7 may be incorporated if desired to obtain fine adjustment of the operating frequency.

Although this circuit is somewhat dependent on the device characteristics which will vary from one sample to

another, it is capable of giving satisfactory performance for most amateur experimenters' requirements. The small size and very low cost make it eligible for substitution into existing equipment using other types of demodulation.

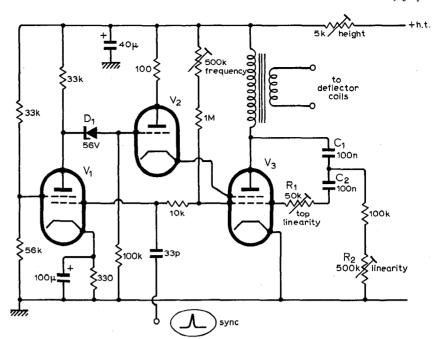
Rodney King, Hastings, Sussex.



Sanatron current timebase

The circuit is a modification of the wellknown Sanatron circuit for use as a current timebase for magnetic deflection. V_3 is the Miller run-down valve, gating being carried out by V_1 and V_2 on the screen grid. To correct for the inadequacies of the output transformer the charging capacitor is split into two halves C_1 and \hat{C}_2 and two resistors R_1 and R_2 are used in a network to provide an approximately parabolic voltage on the grid of V_3 to correct for the differentiating action of the transformer. The gating pulse from V_1 is applied to the screen grid of V_3 via a cathode follower V_2 to minimize loading and allow the use of valves with high screen current for V_3 . The 56-volt zener diode D_1 ensures that V_2 and V_3 are completely cut off when V_1 conducts. Sync can be applied to the grid of V_1 as shown.

The circuit values are appropriate for use as a television field timebase operating at 50Hz. The output transformer should be so chosen to present a load of about $5k\Omega$ to the anode of V_3 when loaded by the scan coils. Different transformers may need different values of feedback components C_1 , C_2 , R_1 , R_2 . Almost any valves can be used for V_1 , V_2 , V_3 ; a 6U8 triode-pentode has normally been used as V_1 and V_2 . Any small output tetrode or



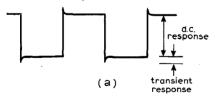
pentode will do for V_3 , the only proviso being that the grid base of V_3 must be longer than that of V_1 , so that V_1 is cut off when V_3 is normally conducting on the rundown. A negative pulse suitable for flyback blanking is available from V_1 anode. None of the resistors in the circuit is suitable for use as a height control, but as the frequency is not very dependent on the total h.t. voltage, variation of this provides a suitable means of controlling amplitude, especially as the current consumption of the circuit is very low.

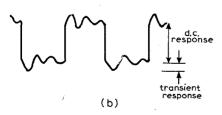
G. Trice, Halstead, Essex.

Transient response testing

The accompanying circuit offers a simple means of testing transient response in regulated power supplies.

The multivibrator is modified by the addition of a diode and a second load resistor on one side to isolate the compound emitter follower stage from the capacitor charging voltage and thus give a sharp front to the waveform. Two series diodes in the base circuits protect the transistors from excessive voltage swings in the switching cycle. The $1k\Omega$ resistor and



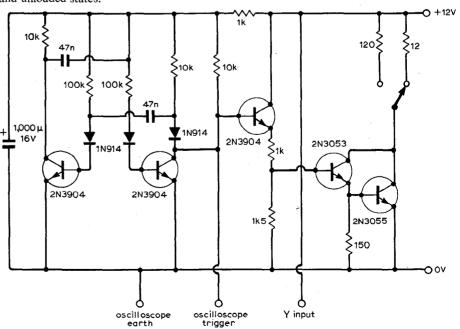


Any tendency to instability degrades a good transient response (a) to that of (b).

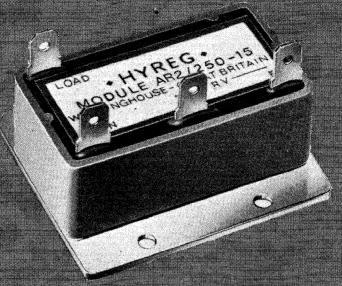
 $1,000\mu$ F capacitor decouple any transients produced by the power supply. The square wave is also taken out to a terminal for oscilloscope synchronization. An emitter follower stage switches the selected load resistor across the output. The transients generated by the supply may be observed on an a.c.-coupled oscilloscope. A good transient response will be similar to that shown in (a), whilst any tendency towards instability would give a response more like that of (b). The d.c. response is the output voltage change between loaded and unloaded states.

Because the transistors are used in a switching mode the dissipation is low, a 2-in square piece of 16 s.w.g. aluminium being sufficient heatsink for the 2N3055, no heatsink being required for the 2N3053. With the values given a 12V supply may be tested at either 100mA or 1A loading. Other voltages and/or currents may be catered for by altering the load resistors.

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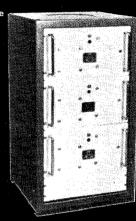
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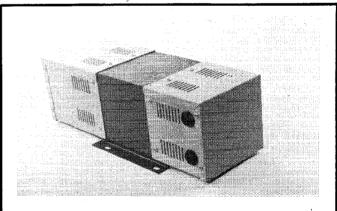
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A.C. Voltage Regulators

A discussion of various techniques for producing a constant a.c. supply voltage

by R. Thompson

This article looks at some methods of regulating a.c. supplies; in particular it addresses the problem of producing a constant level output at nominal value from a 50 Hz supply capable of varying ± 20% from the nominal. The methods all involve the use of thyristors (or triacs) in what may be called synchronous switching. Other forms of regulators have been, and are, used. These include motor-driven variacs, contactor transformer tap changing and ferro-resonant transformers. While these may have advantages for particular applications, they cannot compete in many instances with the speed, accuracy and low cost of the forms of regulator described

The most common form of thyristor circuit for a.c. voltage control is shown in Fig.1, (the thyristors could of course be replaced by a triac). This has been widely used in applications such as lamp dimmers, where a wide control range is required for loads tolerant to high levels of distortion. The form of the load voltage raises a point to be considered in regulating a.c. voltage. What parameter of the output voltage is to remain constant? Normally this will be either its r.m.s. or mean value depending on the characteristics of the load. In some particular instances it may be required to regulate the peak value. In discussing the regulators it is assumed that the r.m.s. value is held constant, the implications of operating mean or peak sensing control circuits will generally be obvious.

The simple phase control circuit of Fig.1 will not meet the requirement of producing a nominal output level. This will require the introduction of an autotransformer as shown in Fig.2. For an r.m.s. output voltage equal to the nominal supply voltage (r.m.s. value of V_o the thyristor firings must be delayed by about 95° when the supply is at its maximum (1.2 V_o). This delay reduces to zero if the supply is reduced to its lower limit (0.8 V_o).

The current rating of the thyristors must be adequate to carry the full load current under delayed firing conditions. Their voltage rating must be in excess of $\sqrt{2} \times 1.5V_o$. This follows from the requirement to delay firing by more than 90°. When the supply is at 1.2 times its nominal value the auto-transformer steps the voltage up to $1.5V_o$ and the thyristors

must withstand the peak value of this voltage. The transformer must have a winding volt-amp capacity of at least $0.6V_oI_L$ as shown in the following expression:

Winding VA rating is given by

AB: $1 \times 0.3 \ V_o I_L = 0.3 \ V_o I_L$ BC: $0.25 \times 1.2 \ V_o I_L = 0.3 \ V_o I_L$ Total VA rating = $0.6 \ V_o I_L$

This form of regulator introduces considerable distortion in the output voltage. Under the worst condition, when the supply is at $1.2V_o$ and thyristor firing is held back by 95°, the total distortion is over 50%. Even if the regulation range of the regulator were reduced to \pm 5% the distortion would still be over 20%.

Trigger circuits for this regulator, when feeding resistive loads, can use any of the types which have been developed for phase control of thyristors and triacs in lamp dimmers and similar applications. The only difference required is that the control of the firing angle must be operated by a circuit measuring the output voltage. Regulating the "mean" output of the regulator is straightforward since a simple mean detecting rectifier and smoothing circuit can be used. Regulating the r.m.s. output is more difficult, particularly with this type of regulator circuit where there is a large change in the form factor of the output waveform. The form factor increases as the input supply increases and the thyristor firing is retarded. This allows a simple form of compensation to be used, as illustrated in Fig.4. If the output r.m.s. voltage is held constant then the mean and peak values of the output voltage vary as shown. A combination of the mean and peak voltages can therefore generate a resultant which can be used to give quite accurate regulation of the r.m.s. output voltage.

The effect of reactive loads will be to modify the conduction periods of the thyristors. For instance, with inductive loads the thyristors will remain conducting after the time of zero crossing of the supply. The change in load voltage waveform is unlikely to prove significant since the first 45° of control has little effect on the r.m.s. output. A more important problem is the fact that the thyristor trigger pulse may occur before conduction from the previous cycle has ceased. This could cause inter-

mittent firing of the thyristors and a limit on the minimum firing angle or prolonged triggering pulses may have to be adopted.

Two-tap regulators

The use of two taps on the auto transformer as shown in Fig. 4 results in very significant improvements in regulator performance. Its operation when feeding

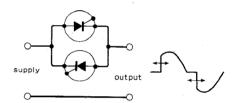


Fig. 1. Simple thyristor phase control and output waveform.

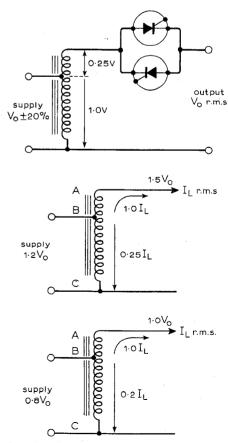


Fig. 2. Single tap regulator. Refer to text for VA rating formulae.

resistive loads is quite simple. At the start of each half cycle one of the lower thyristors conducts, supplying voltage to the load. Some time later in the half cycle one of the upper thyristors is triggered. This increases the voltage on the load and by so doing commutates the lower thyristor. At the end of the half cycle the upper thyristor is extinguished.

It can be seen that the output voltage can be continuously varied between the two values available at the transformer taps by altering the time of triggering the upper thyristors. The distortion is considerably lower than the single tap circuit, being zero at both extremes of input voltage and about 15% for the worst condition. There will be a proportionally lower distortion if the

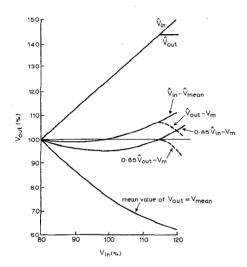


Fig. 3. The characteristics of various methods of generation of the r.m.s. control voltage. The curves are drawn for V_{out} (r.m.s.) constant.

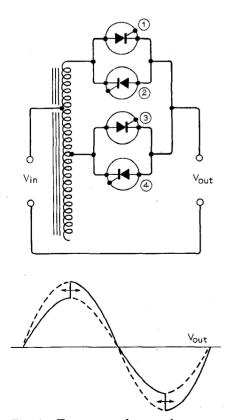


Fig. 4. Two tap regulator and output voltage waveform.

regulation range is reduced below ± 20%. The circuit is still quite simple and because only a fraction of the output power is handled directly by the circuit, ratings of the components are relatively low. Both pairs of thyristors must be rated for continuous full load current since the supply could be at either extreme for long periods. However they do not require as great a voltage rating as those in the single tap circuit. As the load is always connected to one or other of the transformer taps the maximum voltage seen by the thyristors is 0.5 times the nominal peak supply voltage. This is a third of that required for the single tap circuit. The transformer ratings (derived in a manner similar to that shown in Fig.2) are required to be about 25% higher than the single tap transformer, but still low in relation to the load power.

Firing requirements for the thyristors on the upper tap are basically the same as those suitable for single tap regulators. The lower thyristors require prolonged triggering in order to ensure that there is always one or other of the lower thyristors capable of conducting. Provided this prolonged triggering is generated immediately the supply is switched on only the tap-to-tap voltage will ever appear across any of the thyristors. The prolonged triggering can be obtained simply from the trigger windings added to the auto-transformer. A diode and zener clipper can be used for each thyristor as shown in Fig.5.

The problems of reactive loads poses a particular problem for the two tap regulator. This is because of possibility of causing tap-to-tap short circuits. For instance if the load is very inductive and the input supply low, the triggering of the upper thyristors may occur before the current has reversed from the previous half cycle. The first time this occurs, current will flow only in the lower thyristor for the half cycle. If the current is still flowing in the lower thyristor at the time of triggering the upper thyristor in the next half cycle a tap-to-tap short circuit will occur. Fig.6 illustrates this point.

A typical two tap regulator circuit is shown in Fig.7. The lower thyristors are triggered via windings on the autotransformer as explained previously. The thyristors are simultaneously triggered by a common pulse generated by transistor Tr_3 . This transistor is driven by the Shockley diode timing circuits whose charging rate is controlled by the differential amplifier formed by Tr_1 , Tr_2 . At each supply voltage zero crossing the output voltage of the diode bridge (D_{I-4}) falls near to zero. When this happens C_1 is discharged through D_5 and R_1 . This action resets the timing circuit every half cycle of the supply. The purpose of the zener diode in series with the Shockley diode is to ensure that the Shockley is extinguished at this time. When D_5 ceases to conduct, C_1 starts to charge at a rate determined by the collector current of Tr_2 . The Shockley eventually breaks down and thyristors 3 and 4 are fired by the pulse which is generated. At the end of the pulse the Shockley may or may not cease to

conduct depending on whether the collector current of Tr_2 is less or more than the diode holding current. If it does cease to conduct it may go on to produce further pulses during the same half cycle, but these will not produce any consequent change in the thyristor power circuit. The time in each half cycle when the trigger pulse is produced is determined by Tr_2 collector current. This is controlled by comparing a reference voltage, provided by D_6 , with the fraction of the rectified output of the regulator which appears at the base of Tr_1 . If the output voltage increases, the current in Tr_2 will be decreased. As a result the trigger pulse will move back in the half cycle, tending to reduce the regulator output. The sum of the currents taken by Tr_1 and Tr_2 remains substantially constant under all conditions and is determined primarily by the common emitter resistor and the voltage of D_6 . The common emitter resistor, R_2 , therefore provides a convenient method of defining a maximum collector current of Tr_2 , this flowing when Tr_1 is cut off. The resistor is selected to ensure that C_1 cannot charge to the triggering level before a specified time in each half cycle. This "dead time" must be greater than the maximum time taken for the regulator load current to reverse after a voltage zero crossing. The minimum firing angle of the circuit can thus be made always greater than the lagging phase angle of the load.

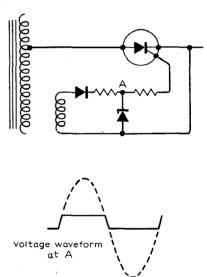


Fig. 5. Thyristor drive circuit.

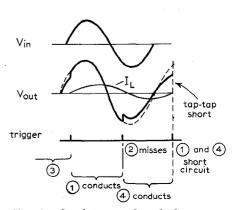


Fig. 6. Conditions under which a tap to tap short circuit can occur.

Fig. 7. Two-tap regulator circuit.

This avoids getting into a condition which could produce a tap-to-tap short circuit.

The resistor R_3 in the emitter of Tr_1 controls the regulator loop gain. This gain must be high for good regulation but loop stability consideration imposes a practical limit. The loop has a substantial lag due to the smoothing of the rectified feedback signal and in addition there is the transport delay introduced by the thyristors.

The thyristors are protected by two fast-acting fuses. Two fuses are necessary because of the possibility of tap-to-tap short circuits. If a single primary fuse were used it would be rated at approximately the same current as the secondary fuses. However, because the current in the primary is lower than in the secondary with tap-totap short circuits, the thyristors could experience a much higher I^2t surge. A "voltrap" suppressor is connected across the transformer taps to limit voltage transients which may occur on breaking the supply to the transformer. The suppressor will also limit voltage transients which occur in the supply voltage. A resistor capacitor network is connected across the thyristors to limit the rate of rise of forward voltage. High rates of rise may be caused by supply transients and they also occur across the lower tap thyristors due to the switching action of the regulator. If the rate of rise of forward voltage is not restricted it may cause spurious triggering of the thyristors.

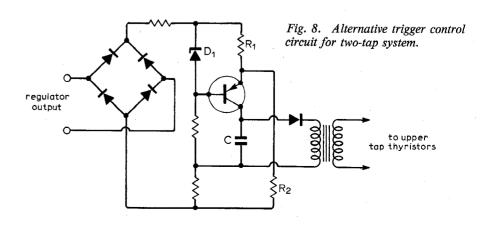
An alternative form of trigger control circuit which can be used for the upper thyristors in the two-tap regulator is shown in Fig.8.

The diode bridge rectifies the output of the regulator and except for brief periods around voltage zero crossing the voltage across R_I is defined by the zener diode voltage. The current through R_I is therefore substantially constant. The current

through R_2 is approximately proportional to the output voltage of the diode bridge and hence to the regulator output voltage. The transistor collector current is thus the difference between the "reference" current in R, and a current proportional to the regulator output voltage. The pulse circuit formed by C and the Shockley diode free-runs at a frequency set by the charging current. If it is initially assumed that this frequency is exactly equal to twice the supply frequency, a consideration of the circuit action will show how this exact condition is maintained. At some time during a half cycle the Shockley discharges C and triggers an upper thyristor. The Shockley extinguishes and the capacitor commences charging. As no smoothing is provided after the rectifier, the current through R_2 , and hence the transistor, will vary considerably during the half cycle. However, capacitor C integrates this varying current and reaches the Shocklev breakdown voltage in exactly a half cycle. By its process of integration and discharge, C averages each separate half cycle of output voltage. The exact frequency required of the oscillations can only be

maintained for an exact average output voltage. If the average output voltage falls, the average charging current of C will increase causing the frequency of the capacitor-diode pulse circuit to increase. This will bring forward the firing of the upper thyristors and hence reduce the output error. As the error reduces, the pulse frequency returns to its original value. Under steady state conditions the capacitor-diode circuit is thus forced to run at exactly twice supply frequency and no permanent error is required at the regulator output. The loop gain of the system does not control the output error but merely the rate at which the circuit can correct output errors. On a half cycle to half cycle basis the gain must be less than unity, that is a given error in one half cycle must not produce overcompensation on the next, for the system to be stable.

The prolonged triggering requirement of the lower tap thyristors will normally make it inconvenient to use a triac in place of these thyristors. Prolonged triggering applied to a triac on the lower tap would lead to a tap-to-tap short circuit. It will normally be simpler to use two thyristors



and only use a triac on the upper tap.

It is possible to alter the configuration of the two-tap regulator in various ways. For instance, the thyristors can be transferred to the primary of the transformer. It would be worthwhile doing this where the load operates at voltages substantially different to those of the supply. For instance the load may be heaters operating at a fraction of the supply voltage. In these situations the auto-transformer function can be included in the main transformer. It should be remembered, however, that the thyristors must withstand the transformer inrush current.

Another alternative is to rearrange the circuit so that the thyristors operate under high voltage/low current conditions. This is illustrated in Fig.9. The general operation and waveforms are identical to that for the circuit of Fig.4. However, the tap-to-tap voltage seen by the thyristors is now higher and the current lower. There are two problems with this configuration. One is that a short-circuit load will produce very high voltages across the thyristors before any fuses blow. The other is the fact that under "no load" conditions the thyristors handle the magnetizing current of the transformer. Since this could be highly inductive it could cause triggering difficulties.

Balanced 3-tap regulators

A regulator circuit which can operate over a very wide range of power factor loads is shown in Fig.10. A centre-tapped inductor has been included in the output circuit of the thyristors. This inductor is connected between the thyristors which could otherwise cause tap-to-tap short circuits due to

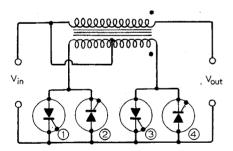


Fig. 9. Circuit arrangement to operate thyristors under high voltage and low current conditions.

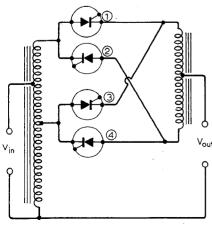


Fig. 10. Balanced three-tap regulator circuit.

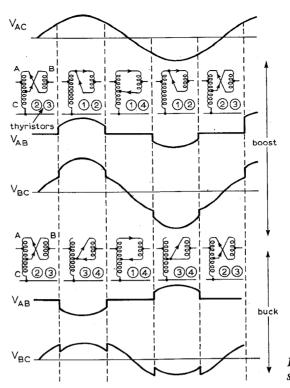


Fig. 11. Regulator switching sequence

faulty triggering on very reactive loads. Apart from eliminating this particular problem, the inductor also acts as an autotransformer and effectively provides a third tap on the transformer balanced between the other two. In order that the inductor should present no significant impedance in series with the supply, the current taken by the load must always be able to split evenly between the two halves of the coil. The circuit can satisfy this condition with only four thyristors if a direct current is maintained through the inductor. This current acts as a bias current through the conducting thyristors and in effect allows the load current to flow in either direction through them. Provided the direct current is greater than half the peak load current the load will never determine the time at which a thyristor is extinguished. It will be shown later that control of the circulating current and control of the output voltage are not mutually exclusive.

Fig.11 illustrates the basic switching sequence of the regulator. The simplified circuits show how the inductor is connected at various times in the cycle (the numbers refer to which thyristors of Fig.10 are conducting). For some time around the centre of each half cycle, the inductor is short circuited and connects the load to the potential of the upper transformer tap, for boosting, or the lower tap for bucking. When the inductor is connected across the taps it behaves as an autotransformer and the output potential is the mean of the two tap voltages. This is equal to the supply voltage if the supply is connected to the transformer midway between the taps. A comparison of the switching sequence shown in Fig.11, with the circuit diagram of Fig.10, will show that commutation from one thyristor to another is obtained simply by triggering the thyristor which must be brought into conduction. For instance, the first transition shown in the boost sequence in Fig.11 involves turning on thyristor 1 and turning off 3. Triggering 1, which will be forward biased at this time, will result in reverse biasing 3. Similar action occurs at all transition in either boost or buck sequences and no additional components are necessary for commutation. It will also be seen that the difference between boosting and bucking is whether 2 is fired before or after 3 on negative half cycles and 1 before or after 4 on positive half cycles. Thyristors 1 and 2 cause transitions in the positive and negative half cycles respectively at similar points in the waveform. Thyristor 2 is already conducting (boost sequence) or reverse biased (buck sequence) when 1 is triggered. In the same way 1 is either conducting or reverse biased when 2 is triggered. Consequently, a common trigger pulse repeating every half cycle may be used for both thyristors. Similar common triggering may be used for thyristors 3 and 4.

The control of output voltage and circulated current is illustrated in Fig.12. The waveforms in (a) show typical boost and buck cycles and it can be seen that the inductor voltage has the same form for either of these conditions. The mean value of this voltage and the inductor winding resistance determine the circulated current. By moving forward both transitions in every half cycle the mean voltage across the inductor is increased, thus increasing the current. This action is shown in (b) and it can be seen that the output voltage is not substantially affected by this change. (The change in waveform symmetry has been exaggerated.) Fig.12 (c) shows the control of output voltage. For this, the two transitions are moved symmetrically about the centre of the half cycle and so do not greatly alter the mean voltage across the inductor. Circulating current and output voltage may thus be regulated with reasonably low interaction between their control loops.

When the supply voltage is near its nominal value, the two transitions coincide and little distortion will be added to the output waveform. The exact voltage at which this happens is determined by the voltage range to be regulated. As the supply voltage decreases, transitions must move further to compensate given changes in voltage. A regulator giving minimum distortion with nominal input will control larger changes in supply voltage above nominal than it can below nominal. Fig.13 shows distortion measured at the output of balanced 3-tap and 2-tap regulators (13(b) and 13(a) respectively) designed to regulate over similar voltage ranges. It can be seen that the regulation range is substantial and distortion would obviously decrease if the regulators were designed for smaller variations in supply voltage.

While the mean voltage across the inductor must be maintained reasonably constant, the alternating voltage across it varies considerably with changes in supply voltage. With a nominal input level, the inductor is continuously connected across the supply, its connections being reversed every half cycle. When the supply voltage is at either extreme of the regulator control range, the inductor will be short-circuited for a large part of each half cycle. The ripple current in the inductor will therefore alter considerably with the supply voltage, and the control circuit must ensure that the minimum instantaneous value of current is always greater than one half the peak load current. If the inductance is chosen on the basis of minimizing the peak energy in the inductor, its value is given by: $L = \frac{2V}{\Omega I_L}$ where I_L is the peak load current,

V is the peak voltage across the inductor and Ω is the supply radian frequency. As the load current should always divide

inductor current voltage (a)

(b)

boost

buck

Fig. 12. Control cycle of voltage and current.

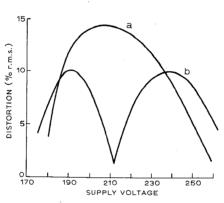
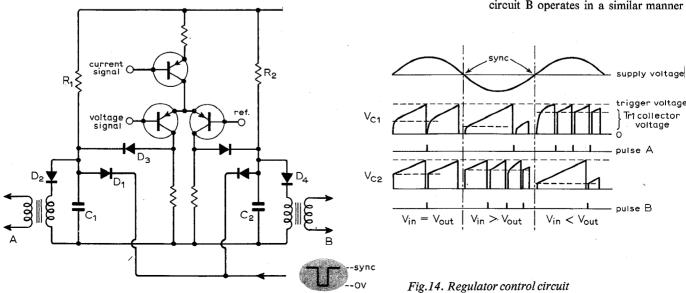


Fig. 13. Output distortion characteristic.

between the halves of the inductor winding, it will not contribute to the energy stored in this element.

The balanced 3-tap circuit provides full regulation for any load power factor and produces an output waveform substantially better than that obtained by a 2-tap regulator. The price paid for this improved performance is a reduction in the load voltamp rating for a given thyristor and transformer cost, (or alternatively a reduced regulation range). The effect on the ratings depends on the exact nature of the load to be accommodated and the design adopted for the inductance. For a given cost, the balanced 3-tap circuit will handle about half the load volt-amps that the 2-tap circuit will.

A control circuit for the balanced 3-tap regulator must be capable of generating two separate trigger pulses in a sequence repeating every half cycle. (It was explained earlier that thyristors may be triggered in pairs.) These two pulses must move symmetrically about the centre of each half cycle to control output voltage and move asymmetrically to control circulated current. The elements of a circuit which provides this action is shown in Fig.14. The two pulses are produced by Shockley diodes D_2 and D_4 discharging ramp generators C_1R_1 and C_2R_2 . The time taken for a capacitor voltage to reach the breakdown level of its associated Shockley diode is set by the time constant and aiming potential of the CR circuit and by the level to which the capacitor is reset after each pulse. Each time that C_1 is discharged, diode D_3 conducts and rapidly resets the capacitor to the voltage level of Tr_1 collector. D, then turns off and the capacitor voltage continues to rise at a rate set by the current through R_1 . A synchronizing pulse occurring at supply voltage zero crossings discharges C_1 to zero each half cycle via diode D_1 . The collector voltage of Tr_1 , by resetting C_1 through D_3 , controls the delay between the end of the synchronizing pulse and the generation of an output pulse at A. Further pulses which may be produced by D_2 before the next synchronizing pulse will not affect the regulator action since they cannot cause the firing of any thyristor. It can be seen that pulse circuit B operates in a similar manner and



is controlled by Tr_2 collector voltage. The circuit waveforms for the condition V_{in} = V_{out} show that Tr_1 and Tr_2 collectors reset C_1 and C_2 to the same level and pulses at A and B both occur at the centre of each half cycle. If the "voltage" signal to the differential amplifier is altered, the voltages of the two collectors move in opposite directions causing the pulses at A and B to move in opposite directions. For $V_{in} > V_{out}$ pulse B leads A and for $V_{in} < V_{out}$, A leads B. Changes in the "voltage" signal therefore produce the symmetrical movement of pulses which is required to control the regulator output voltage. By varying the "current" signal, the amplifier tail current is altered and the two collector voltages are made to change in the same direction. This causes the two pulses to move in the same direction and alter the circulating current by unbalancing the output waveform. If the current control signal is derived by directly measuring inductor current, both halves of the winding must be monitored so that voltages due to load current can be cancelled. Alternatively, a signal may be derived from the magnetic field of the inductor which will be independent of load current.

General points on thyristor regulators

The characteristics of thyristors and triacs are normally well described in manufacturers' data. However it is worth noting a a few general points with regard to their use in a.c. voltage regulators.

High rates of rise of current through the thyristors during the time that they are being brought into conduction can lead to device failure. In the regulator circuits which have been described the leakage inductance of the transformer will normally provide adequate protection. This should be checked and it is also important to ensure a high level of trigger current during this switching time.

High rates of change of voltage across the thyristors can cause spurious triggering. Step changes of voltage occur across the devices due to the normal operation of the regulator and in addition transients on the supply can also cause sharp changes. Because of this it is advisable to shunt the thyristors by RC circuits as shown earlier in Fig.7. It should be remembered that such shunting increases the initial rate of rise of current through the thyristors.

The shunting of the thyristors by RC circuits is beneficial in two other respects. Some protection is provided against excessive voltages appearing across the thyristors due to short duration transients such as those caused by switching of the supply or load. Also the slowing of the change in voltage provides an output waveform with a much lower level of radio frequency harmonics.

The difficulty associated with radio frequency interference will of course depend on the application. A good metallic screen for the regulator housing with a balanced choke-capacitor circuit on the output will be adequate for most applications. One thyristor manufacturer

recommends that for domestic situations radiated interference will be satisfactory if the rate of change of current in the supply wires is kept below 0.35 A/ μ s. This will normally require an inductance of a few hundred microhenries.

Finally it is worth making some comparisons between thyristor regulators and the most commonly used alternative, the ferro-resonant constant voltage transformer. Such a comparison can only be made in very approximate terms since aspects such as cost will be highly dependent on the precise requirements - see Tables below. The cost quoted for the thyristor regulator represents a likely production cost of a 2-tap circuit. It should also be remembered that although both forms of regulator carry out a similar task, that of providing a stabilized a.c. voltage, there are important differences in their characteristics. In particular circumstances these detailed differences may be very important and override other considerations. For instance where it is important to minimize weight, the thyristor regulator is an almost automatic choice. Where frequent load short circuits occur the ferro-resonant transformer may have to be used.

TABLE 1
Performance characteristics

| | | F |
|--|------------------------|----------------------------------|
| Characteristic | Thyristor Regulator | Ferro-resonant transformer |
| Input voltage | | . 150/ |
| range | ± 20% | ± 15% |
| Stability of output voltage (for full change of supply voltage and load current) | ± 2% | ± 3% |
| Overload protection | Fuse | Can handle continuous short |
| Output distortion | < 15% | circuit. Approximately 20% |

TABLE 2
Weight and cost comparisons

| Rating | | Thyristor Regulator | | Ferro-resonant transformer | |
|---------|--------|------------------------|---------|-------------------------------|--|
| | Weight | Cost | Weight | Cost | |
| 500 VA | 14 lbs | £25 | 45 lbs | £33 | |
| 1000 VA | 20 lbs | £35 | 85 ibs | £65 | |
| 1500 VA | 28 ibs | £50 | 120 lbs | £100 | |

Books Received

Nomographs for Electronics — Instant calculations for designers by Robert L. Peters, Ph.D. is a collection of nomographs covering areas from basic electrical circuitry to supporting systems components such as motors and generators. The nomographs are for use by engineers, students and technicians to determine the characteristics of electronic circuits and the values of their components. Before presenting the nomographs for each subject area, the author provides a summary of the theories involved, followed by instructions explaining how each nomograph is used. The subject areas covered include a.c. and d.c. networks, semiconductor devices, valves, filters, relays, passive devices, transformers, generators, servosystems, instrumentation, and motors. aerials. Most of the nomographs are of the simple linear-alignment variety to increase accuracy and reduce the chance of error. Price £11.25 (incl. p/p). Pp.276. Cheques and money orders payable to Business Books Centre Ltd, should be sent to Wireless World, Room 23, Dorset House, Stamford Street, London SE1 9LU.

Simplified Statistical Analysis — handbook of methods, examples and tables by Harry H. Holscher is designed to help the reader get the maximum amount of information from a minimum amount of experiment and effort. It should be particularly useful for engineers, managers, scientists, teachers or social scientists who need to do statistical analysis. The author provides methods of analysis, 107 examples, 122 tables and 39 graphs. In the case of worked examples, graphic or tabular visual presentation is used in preference to complex mathematics. Chapter include: the principle of randomization; planning of experiments; analysis of experimental data and comparison of sets of data for significant differences or for precision in measurement. Price £6.20 (incl.p/p). Pp.235. Cheques and money orders, payable to Business Books Centre Ltd, should be sent to Wireless World, Room 23, Dorset House, Stamford Street, London SE1 9LU.

Television Engineers' Pocket Book by P. J. McGoldrick (6th edition) has been necessitated by the dramatic change that has taken place in broadcasting following the introduction of colour transmissions. It is hoped that this edition will provide all the information that the TV engineer requires, whether he is servicing a monochrome or a colour receiver. In addition to the general revision and updating of the book, a considerable amount of new material has been added particularly on the subject of receiver circuitry and colour, integrated circuits and servicing. The aim has been to provide all concerned with, or interested in, the installation and servicing of modern television receivers with a useful summary of basic facts, circuit techniques and technical data for everyday reference. Price £2.50. Pp.372. Butterworth & Co. Ltd, 88 Kingsway, London WC2B 6AB.

Beginners' Guide to Colour Television (2nd edition) by Gordon J. King has been completely rewritten by the author and will be of value to all those who wish to understand how and why a colour TV works. Chapter headings include: historical outline; colours and signals; pictures and signals; complete signal; PAL receiver; SECAM basics. Price £1.95. Pp.189. Butterworth & Co. Ltd, 88 Kingsway, London WC2B 6AB.

Circards — 8

Astable circuits

by J. Carruthers, J. H. Evans, J. Kinsler & P. Williams*

Many circuits generating periodic rectangular waveforms depend on changing the charge on a capacitor. This rate of change of charge is frequently determined by an *RC* circuit, which inherently produces exponential waveforms when connected to a voltage source (Fig. 1). Such waveforms may be used to control the switching on and/or off instants of an active device.

An astable multivibrator using discrete components is a commonly occurring example of a circuit employing this principle (Fig. 2). This uses two transistors, cross-coupled via capacitors C_1 and C_2 . Normally this circuit has no stable states (they may be called quasi-stable states), but switches alternately from the one state of Tr_1 saturated, Tr_2 off, to the state of Tr_2 saturated, Tr_1 off. For a given value of V_{CC} , the rate of switching depends on the time constant C_1R_4 , C_2R_3 , and on the potential to which the base resistors R_3 and R_4 are returned. The more conventional configuration will return R_3 and R_4 to the $+ V_{CC}$ rail to give a period independent of rail voltage.

Fig. 3 shows that the output rectangular waves available at the collectors are in anti-phase, and their mark-to-space ratio may be varied by adjusting the C_1R_4 and C_2R_3 time constants. Some astables are attractive due to the small number of components required, but they might be considered to have certain disadvantages. The use of a single CR network gives an output waveform having a non-unity mark-to-space ratio which may be difficult to closely control. Also a second anti-phase output is not available. A circuit using a single capacitor that overcomes these objections to an extent is the emittercoupled astable of Fig. 4. In this arrangement, two emitter resistors are used to allow independent adjustment of the mark and space times of the two output waveforms.

Another group of astable circuits apparently different from each other fit into the form of the general bridge network shown in Fig. 5. The amplifier block is provided with two external networks. One network, providing negative feedback, ensures that the d.c. conditions in the amplifier are such that it sits near the middle of its operating region

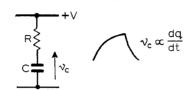


Fig. 1. Series RC produces an exponential waveform across the capacitor.

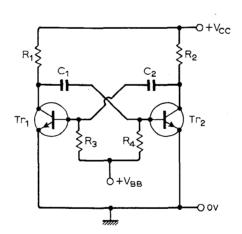


Fig. 2. Basic astable multivibrator. Variation of V_{BB} changes the period.

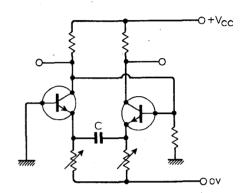


Fig. 4. A single-capacitor astable multivibrator providing mark-to-space ratio adjustment and antiphase outputs. Circuits using one capacitor are normally less flexible.

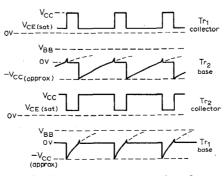


Fig. 3. Waveforms encountered in the circuit of Fig. 2, where C_1R_4 is longer than C_2R_3 .

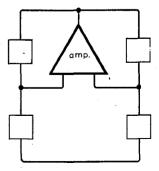


Fig. 5. Basic diagram of the group of oscillators using both positive and negative feedback. The circuit is seen to be a bridge network with a sustaining amplifier.

^{*} All at Paisley College of Technology.

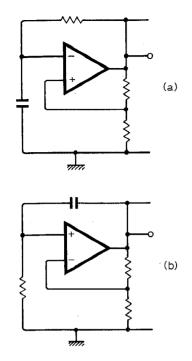


Fig. 6. Two possible combinations of feedback types using a single differential amplifier.

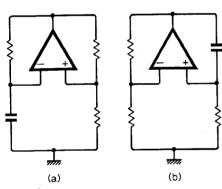


Fig. 8. Circuits of Fig. 6 redrawn to show that they conform to the general bridge circuit of Fig. 5.

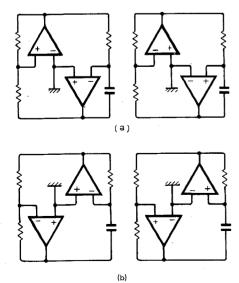
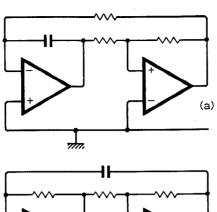
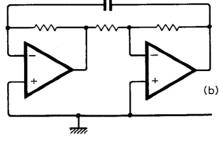
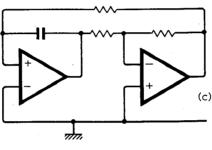


Fig. 9. The four twin-amplifier circuits are also examples of the general bridge circuit.







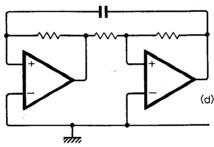


Fig. 7. Four combinations of RC feedback and amplifiers used in the single-ended mode.

where the gain is high. The other network, which provides positive feedback, forces the amplifier to switch between two distinct states. This amplifier block may comprise either a single differential-input amplifier or two single-ended types. In conjunction with the different combinations of feedback components, this leads to the apparently different configurations shown in Figs. 6 and 7. These circuits have been redrawn in Figs. 8 and 9 respectively to show their relationship to the bridge arrangement. Circuit 7(d) does not operate as an astable because it will permanently latch into one state, since positive feedback is applied across each single-ended amplifier.

The output waveform from an astable multivibrator is not necessarily rectangular. A self-retriggering action can be obtained by using an electronic switch to discharge a capacitor when its potential exceeds a preset d.c. bias which holds the switch open.

To provide circuits compatible with logic levels, many astables operate in their saturated mode to give well-defined voltage limits. This is achieved at the expense of the switching rate, but may be avoided by the use of current mode switching techniques.

Other applications demand long periods with accurately defined transition times, and this requires high stability passive components. Reasonably small value but stable capacitors can be used in conjunction with field-effect transistors to provide the long time constants necessary. It should be remembered that if an astable multivibrator has good frequency stability, it may prove difficult to synchronize it from an external source.

The reader will have noted that inductive timing elements are conspicuous by their absence. The reason behind this is that, in comparison with capacitors, inductors tend to be more costly and physically larger. One type of inductively coupled circuit worthy of mention is the astable blocking oscillator, which is capable of producing output pulses having a very small mark-to-space ratio.

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The Realm of Microwaves

4. Microwave integrated circuit technology and lumped element design

by M. W. Hosking, * M.Sc.

Five years ago, hybrid microwave circuit design existed mainly within the laboratory and there were very few examples of production-standard circuitry. Now, most companies engaged in microwave subsystem design have experience of microwave integrated circuits and their commercial applications have been, and still are, rapidly increasing. For this to happen, four basic things were needed:

- -an accurate design theory
- -a high-definition pattern technology
- a ready supply of consistently highquality substrate materials
- —availability of semiconductor devices in suitable and well-defined encapsulations. This pre-supposes that sufficient potential exists to make all the effort worthwhile and the incentive has been the saving in size and weight possible with this type of circuitry. In particular, emphasis has been placed on lightweight receiver design for satellite and airborne installations and on the high-volume application in phased-array radar.

When referring to microwave i.cs, the type of transmission line inferred is the microstrip form described in the March and June installments. That is, a high dielectric-constant substrate metallised on one side to form a ground plane and with the microwave components defined by suitably shaped conductor patterns on the other side. The microwave electromagnetic field is largely confined to within the substrate, the properties of which thus become of great interest.

Apart from lumped elements, dealt with later in this article, the passive components used are of the same order of size as the microstrip wavelength. This means that inductances, capacitances and resonant circuits can be produced by discrete lengths of transmission line. Also, abrupt changes in the linear dimensions of the line produce local disturbances of the field and have the effect of reactive elements.

Microwave design theory is thus much more complex than that used at lower frequencies and is particularly so in the microstrip form of circuit. The significant theoretical advances made have now enabled the first-listed condition to be satisfied, especially for the cases of coupled pairs of lines and the deduction of equivalent circuits for various transmission line discontinuities.

Fabrication of microstrip conductor patterns has benefited from mask-making and etching techniques already in existence in the vacuum deposition and i.c. fields. These techniques have been capable of direct application to microwave i.cs with the result that very little development work has been necessary, together with little modification of existing equipment.

The overall requirement is for consistently accurate linewidths and spacings together with fine edge definition. This last condition depends on the metallization thickness, method of production and the surface finish of the substrate. As well as good surface finish, the substrate must also be dimensionally accurate and have a well-defined and consistent dielectric constant together with low loss.

Thanks to the steadily increasing demands for substrates, more suppliers are marketing materials with well-enough controlled properties; whereas previously the circuit designer had to cut, grind and polish his own materials. Finally, demand has tended to create the supply of semiconductor devices in package forms suitable for bonding to the high-dielectric constant substrates. There is now a large variety of styles, well defined as to their parasitic reactances and capable of being handled by existing hybrid equipment and bonding methods.

Skin effect

When a steady current flows through a conductor, it is distributed uniformly throughout the conductor's cross section. However, under alternating conditions, there is a tendency for the current to concentrate near

the surface, resulting in an uneven distribution. Moreover, as the bulk of the current flows in a more restricted space, the effective resistance of the conductor is increased from its d.c. value and keeps on increasing as the frequency is raised.

Starting from the surface, the current density decreases exponentially towards the centre of the conductor and the distance at which its value has decreased to 1/e (e = 2.7182) of its surface value is called the skin depth. At this point, incidentally, its phase lags behind the surface phase by one radian.

Skin depth is inversely proportional to the square root of both frequency and conductivity and can be very small in the microwave region, resulting in high current densities and high resistive losses. These losses become severe if the conductor thickness is less than the skin depth and conductor patterns on microstrip line are usually made four to five times the skin depth. Fig. 1 shows the relationship between skin depth and frequency for some commonly-used microwave conductors, based on the equation

skin depth,
$$d = \left(\frac{1}{\pi f \sigma \mu}\right)^{\frac{1}{2}}$$

where f is frequency, σ conductivity and μ the absolute permeability of the materials given by $\mu = \mu_o \mu_r$. To compare conductor losses, the surface resistivity R_s is often quoted and this is the resistance of a square section of conductor one skin depth thick.

Why should the current concentrate near the surface, just because the frequency is

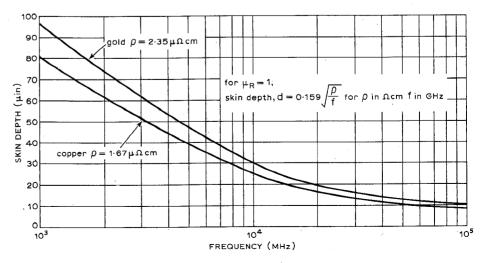


Fig. 1. Skin depth indicates extent to which h.f. current penetrates the conductor. To minimize resistive loss the conductor thickness should be several skin depths.

increased? Taking the flat strip case drawn in Fig. 2, there will be lines of magnetic flux associated with a current flowing in the strip. This gives rise to inductive effects because an inductance is produced by flux-linkages and is directly proportional to the number of such linkages.

Current near the centre of the conductor is encircled by more lines of magnetic flux than is the current near the ends of the strip and so the centre of the strip has a higher inductance than the surface. As the frequency increases, so too does the inductive reactance of the conductor, with the result that the centre has a higher impedance than the sides. So, taking the path of least resistance, the high-frequency current concentrates near the surface; in this example, most current is at the edges.

If the conductor were circular in section instead of a strip, then the current would be evenly distributed around the surface; implying that not only must the electrical properties of a conductor be taken into account, but so must its geometry.

Now you know why a good surface finish is required from a microwave i.c. substrates.

Substrate materials

The number of materials suitable for general microstrip and microwave i.c. usage is really quite limited. To keep circuits small, a high dielectric constant is desirable, but this is kept from being too high by the present capabilities of line-definition technology. For, as the dielectric constant increases, then so must the substrate thickness decrease to prevent the exitation of higher-order modes. This in turn means that, for a given impedance, the linewidth must reduce proportionally, hence the limit.

A low dielectric loss is a further requisite which, together with providing both a good surface finish and good adherence for conductors, severely limits the range of suitable materials. The table lists the four types of substrate generally used, in order of popularity, together with some relevant properties. Alumina has become the standard substrate and can be obtained now from many vendors. Although almost any shape and size can be made, dimensions in multiples of $\frac{1}{2}$ inch and a 0.025 inch thickness have become stock sizes. The alumina can be obtained in a range of purities, typically between 94% and 99.9%, with 99.5% being the preferred compromise between the electrical and mechanical properties.

Sapphire is capable of yielding a better surface finish than alumina, enabling a better line definition to be obtained and giving slightly lower conductor losses. But, apart from its use at the higher frequencies,

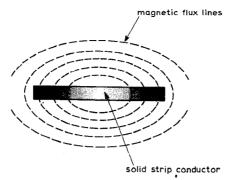


Fig. 2. Current flows nearer to surface because of lower inductive reactance in this region, but conductor geometry plays a part as indicated by the flat strip.

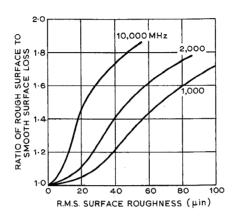


Fig. 3. Surface roughness is important and increases conductor loss. Good-quality microwave i.cs have a surface roughness better than 5µin.

above about 20,000 MHz, where such factors have more significance, it is seldom used in place of alumina for microwave i.cs.

However, for lumped-element circuits, wherein component dimensions tend to be smaller and the detail finer than in microwave i.cs, sapphire is often used for the substrate. In cases where a very thin, say 0.005–0.010 inch, substrate was required, sapphire would be preferred because of the strength afforded by its crystalline structure, as opposed to the sintered structure of alumina. Ferrite substrate is steadily gaining in application as a dual-purpose medium: acting as a normal transmission line and also one whose properties can be locally varied by applying a magnetic field.

In this way, components like isolators, circulators and phase shifters can be produced on a continuous substrate. The widely-used alternative is to cut a hole in the alumina or sapphire circuit and insert a piece of ferrite material to produce such components; but this involves a high degree of mechanical

skill and extra electrical design complexity in compensating for the abrupt change in dielectric constant. Fused quartz is not so widely used as the other materials, partly because of its higher loss, but it finds application when a low dielectric constant is required together with a high-temperature capability. Because of its good surface finish it is also used as a lumped-element substrate. Finally, beryllium oxide is used only when all else has failed and then only for its superior thermal conductivity, which is a factor of 10 better than alumina. This makes it suitable for the heat-sinking of high-power devices—usually transistor chips—but the material needs special handling due to the health hazard from BeO dust.

Conductor technology

Three main properties are required of microwave i.c. conductors:

- -good adhesion to the substrate
- —low resistivity
- -easily bonded to the active devices and there are two different processes used to produce the conductors themselves: thick film and thin film. In the thick film process, the conductor consists of a proprietary mixture of metal powder, glass and binders in ink form which is screen-printed onto the substrate and then fired. From the skineffect loss considerations mentioned previously, the conductor thickness is made at least 4-5 skin depths. It is best to keep the conductor thickness as small as the skin depth limit permits in order to obtain better line definition and freedom from undercutting. Also, once the thickness becomes more than about 1% of the substrate thickness, it starts to noticeably influence the electromagnetic fields and so must be taken into account during circuit design.

The thin film process implies vacuum deposition of the conductor material by such standard methods as evaporation and sputtering. Silver, copper and gold are the most often used metals but, unfortunately, these do not adhere very well to the substrates.

Adhesion can be improved by using a coarser surface finish: giving more of a mechanical bond to the film, but this is not the only consideration. Surface roughness interrupts the conductor skin current and, especially transversely to the current flow, will cause eddy current losses. These can increase the theoretical loss very substantially at microwave frequencies as can be seen from Fig. 3. A 20 microinch finish at 1,000 MHz gives about 5% extra loss, but this has increased to 45% at 10,000 MHz.

This band represents the portion of the spectrum in which most work to date on microwave i.cs has concentrated. So, at the higher end, a surface roughness of 5 microinches or less is required. In addition to the increase in conductor loss, surface roughness sets a limit on the attainable line and spacing definition. Exact correlation is difficult, as processing "technique" plays a part, but one would not expect good linespacing definition below about 10 times the r.m.s. surface roughness.

From these considerations, surface roughness is not used to improve the adhesion of thin films. Instead, a chemical bond can be formed by vacuum depositing

Properties of substrates

| Substrate | Tan δ at 10,000MHz | ε _r at 10,000MHz | Thermal cond (Wcm ⁻¹ ,°C ⁻¹) |
|-----------------|-----------------------|--------------------------------|--|
| alumina | 2×10-4 | 9.7 | 0.2 |
| garnet ferrite | 2.5×10 ⁻⁴ | 15.5-16.5 | 0.03 |
| sapphire | 2×10-4 | 11.0 | 0.48 |
| fused quartz | 1.5×10 ⁻³ | 3.78 | 0.09 |
| beryllium oxide | 10-4 | 6.0 | 2.5 |

the film onto a substrate which has been previously coated with a metal oxide. Of course, this metal in turn must have good adhesion and an eminently suitable one is chromium. This is first deposited as a seeding layer about 200Å thick $(1Å = 10^{-8} \text{ cm})$ onto which is then evaporated the main conductor. It is not generally satisfactory, due to film stresses and definition problems, to build up the full conductor thickness by vacuum deposition. Instead, a nominal thickness, sufficient to cover the peaks in the surface roughness, is deposited and then the remaining material is formed by a normal electroplating process. If the microwave i.c. were for use at 1,000 MHz, with gold as the conductor, then the skin depth is 100 microinch or 1/10 of a thou. This much might be applied by vacuum deposition and subsequently built up to five skin depths, $\frac{1}{2}$ a thou., by plating.

Gold is used as the top metal because of its non-tarnishing properties which, in turn, enables good bonds to be easily made with active devices. Copper and silver, though, have slightly better electrical conductivity, but are easily tarnished and more difficult to bond. So, while the Cr-Au system is most commonly used, slightly lower conductor loss can be obtained by a Cr layer, followed by deposited Cu, plated up with more Cu and topped off with a flash of Au for bonding.

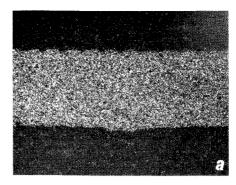
Producing the conductor pattern

Using either the thick or thin film process. there are two methods of defining the actual microwave circuit pattern: it can be formed directly on the surface, or it can be etched out from a completely covered substrate.

The former method involves production of a mask through which the conductor is either screen printed or evaporated and the latter can be carried out with conventional photo-etching techniques. The direct method is more suited to thick films, as the fired inks are more difficult to etch accurately and it is not used very often with vacuum deposition because of the definition problems in building up the thick metal.

In the other, photo-etching process, the required circuit is first of all drawn say 20 times full size using an opaque tape on a transparent background. With care, the dimensional accuracy can be kept better than 0.020in, so that on completion of the next stage, which is photographic reduction to a full-sized negative, the error reduces to 0.001in. By taking additional time, trouble and mechanical aid within a controlled environment, accuracies considerably better than this can be obtained.

The substrate, completely coated with metal, is then covered with photoresist, usually a u.v.-sensitive type, and dried. The negative is then laid on top, emulsion side down to avoid shadowing, and exposed; with the result that the clear areas, representing the conductor pattern, allow the light through to chemically cross-link the resist. These exposed areas are now resistant to solvent attack and remain fixed to the metal while the remainder of the resist is washed off. Thus, in the next stage, which is etching, all of the conducting metal will be removed apart from those areas protected



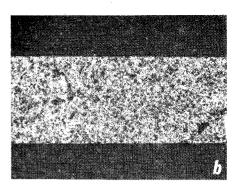
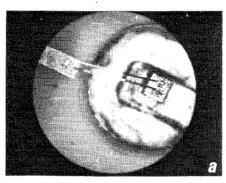


Fig. 4. Comparison of (a) an high-quality thick-film gold line and (b) a thin-film photoetched one, both on alumina substrate. Mesh size limits edge definition of the thick-film line which is about 0.020in wide. The thick-film line was provided by J. R. Corkhill of R.A.E.



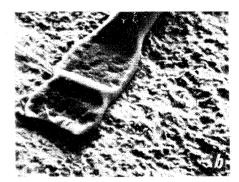


Fig. 5. Photograph is of a 0.001-in dia. aluminium wire, ultrasonically bonded to a semiconductor chip, (a), and an electron microphotograph of a similar bond (b).

by the resist. If it was decided to vacuum deposit the pattern then this process could be used to make the metal mask.

There are, therefore, two techniques which can be used to deposit the conductor pattern and two techniques for etching it away. Of the former, the thick film process has not been so widely used at microwave frequencies as the thin film, but improvements to the technology are steadily making it more popular.

The main disadvantage of the thick film process is the difficulty in obtaining fine edge definition, together with small linewidths and spacings of less than about 0.002in. This is partly due to the slightly porous nature of the fired inks, but mainly because of the finite size of the printing screen mesh. Fig. 4 gives an idea of the difference between a thick and thin film gold line on alumina substrate. It is, however, not unusual to find thick film resistors used with hybrid microwave i.cs either as the d.c. bias resistors for a transistor oscillator or perhaps in the i.f. circuitry of a receiver.

The thin film vacuum deposition process is at present the most popular one and there are the two methods open of forming the actual pattern. Of these, the photo-etching process is the one most widely used for the following reasons: during prototype development, having to re-design a metal mask is more costly and inconvenient than producing another photograph; finer definition can be achieved with etching; definition can be lost during the plating-up stage of a mask-deposited pattern.

In summary, microwave i.cs are formed on the substrates listed in the table by photoetching the conductor patterns from a vacuum deposited film, then plated up to several skin depths in thickness. The difference between low frequency printed circuit design and microwave i.c. design is largely a matter of degree. In the latter, everything is of great importance electrically: the dimensional accuracy and purity of the substrate and conductor as well as dimensional accuracy of the circuit pattern.

Device bonding

In response to the increasing use of hybrid circuits, semiconductor manufacturers have widened their range of encapsulations in which the active devices can be bought. The area in which this was most needed was for package styles suitable for series-mounting with a microstrip line. Shunt-mounting is necessary for good heat-sinking, but involves drilling holes through the very hard substrate and is to be avoided when possible.

Package styles now available vary from just the chip itself, which might be a 0.015in cube with metallization on two sides, to the beam-lead type shown in Fig. 7(b) of Part 3 and on to various low-profile, hermetically sealed packages with flying leads. A widely used method of attaching the active devices to the substrate is thermo-compression bonding and can be used for fixing both the chips and the leads. Once again, this particular technique was already in use with the low frequency i.cs and did not have to be specially developed for microwave i.cs. For

the bonding of a chip, the substrate with the etched-out conductor pattern is heated to about 320°C and the gold-plated side of the chip is pressed down in contact with the gold-plated conductor. Using a contact pressure of a few hundred grams, the two gold surfaces will fuse together to produce a good bond. The holder which carries the chip is often ultrasonically vibrated to "scrub" the two surfaces together and, when this is done, a slightly lower bonding temperature can be used. To attach a wire to the bonded chip, or to attach devices already possessing leads, to the substrate, thermo-compression bonding is again most commonly used. In this case, a probe either heated or unheated is used to press the lead and substrate together. The geometry of the probe tip is an important factor in the bonding process and ultrasonic vibration can be applied to permit the best compromise between bonding area, pressure, temperature and time to be made. An example of wire bonding is given in Fig. 5.

Lumped elements

Readers following this series will recall the general description of distributed microwave circuit elements in Part 3 and it has been the intention, so far, to emphasise the vast difference between the practical realization of a microwave circuit and that of its lower-frequency counterpart. Our earlier definition of the microwave spectrum indicated a wavelength range of 30cm to 1mm; so that, in a conductor system supporting microwave fields, this means that significant changes in phase can occur within quite a small physical distance. Thus, as we have seen, inductive and capacitive effects can be produced from a length of transmission line, as opposed to the discrete (or lumped) elements used at lower frequencies. Take, for example, a length of 50ohm microstrip transmission line on alumina operating at a frequency of 3,000MHz. We have seen that this requires a width to thickness ratio of unity, so that, using the effective dielectric constant given by Fig. 4 of Part 2, the transmission line wavelength is about 4cm. A distance of 1cm, therefore, represents a phase shift of 90°-sufficient to change an inductive reactance into a capacitive reactance. In a conventional electronic circuit though, operating at, say, 300 kHz, the same distance of 1cm might be typical of the size of a resistor or capacitor. But the phase change across this component would be a negligible 0.0036° and would not affect its particular value of R, C or L. A lumped element, therefore, is one whose value does not change significantly with frequency; and what is considered as being significant must depend on the particular application.

Generally, lumped elements are smaller than about 1/40 of the transmission line wavelength. Conventional components start to show distributed effects above a few hundred MHz and they also become very lossy, so that different techniques must be used to produce the circuits. Fortunately, the higher the frequency, the smaller the value of inductance and capacitance necessary to produce a given effect and, at microwave frequencies, one is usually concerned with values of a few nH and fractions of a nH and a few pF and fractions of a pF.

Fabrication of the lumped elements is carried out using the photo-etching process to define the circuits on good-quality substrates. The technology used is most important as, in order to be competitive with distributed circuits, high Q factors and correspondingly low losses must be achieved. Inductors are formed from a length of narrow conductor ribbon as in Fig. 6(a) which, for inductances greater than about 1nH, can be extended into a planar spiral.

Single-layer capacitances can be formed from the etched conductor ribbon by an interdigital arrangement shown in Fig. 6(b), where use is made of the microwave fringing fields. This makes circuit processing simpler but, if a higher capacitance per unit area is required, then a three-layer metal-dielectric/metal sandwich can be used as in Fig. 6(c).

Lumped inductors

Any conductor possesses inductance, the value depending upon its geometry, and a commonly used formula for predicting the inductance of an etched strip is that given by F. E. Terman¹⁵. For a strip whose thickness is very much smaller than its width, the

inductance per unit length is a function of the length to width ratio, l/w. If the operating wavelength were, say, $10 \text{cm} (\sim 4 \text{in})$, and the restriction that a component must be smaller than 1/40 of a wavelength were applied, then the permissible inductor length would be 0.1 in. For a width of 0.010 in the l/w ratio is 10 and, from the formula mentioned, the attainable inductance is 1.8 nH.

If the conductor had been 0.005in wide, the inductance would have been about 2.6nH or, alternatively, the original inductance could have been obtained with a shorter strip. However, another factor must be taken into account before reducing the size too much and that is the inductor Q. This can be represented as the ratio of the reactance to the resistance and this resistance is the high-frequency value, which takes into account the skin effect.

It has already been shown that the skin effect increases the nominal d.c. resistance and is also a function of conductor geometry, the lowest resistance being obtained with a wide strip. In addition, the conductor thickness must be several times the skin depth. Theoretically, the Q factor of a lumped element inductor increases as the square root of frequency and this relationship is achieved in practice for a strip inductor at frequencies well into S-band (2,000-4,000MHz). Thereafter, the rate of increase of Q slows down as, for the same l/w ratio, the inductor becomes less "lumped". Q factors greater than about 50 are required in order to achieve element losses of several tenths of a dB and, by taking care with substrate and conductor, such values can be obtained at 1,000MHz, rising to a maximum of about 100 as the frequency increases.

So far, attention has been confined to the strip inductor but, for higher inductance values, this can be formed as a circular or square spiral such as that shown in Fig. 7. Provided that the outside dimensions meet the size requirements, such an element will still appear as lumped. Its inductance, however, will be less than that of a strip of the same unwound length, due to the mutual proximity of the turns. Once again, the highest Q factor results from wide strips which, for a given inductor size, results in

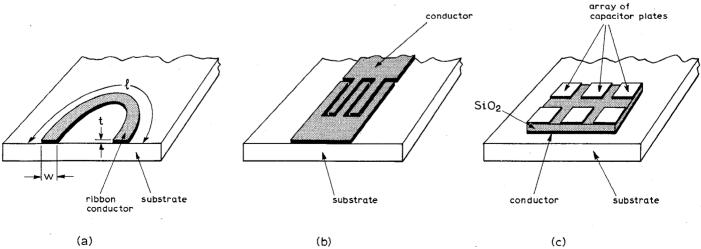


Fig. 6. Types of lumped element component: (a) simple strip inductor; (b) interdigital capacitor; (c) parallel plate capacitor. The inductance of (a) is proportional to its length and the Q factor to the width/length ratio. Additional capacitance can be added to (b) by forming a ground plane on the other side of the substrate, while a degree of tuning can be obtained from (c) by linking numbers of pads together.

small, inter-turn spacings of the order of 1 or 2 thousandths of an inch. Typically, circuits at L band (1,000-2,000MHz) or S band (2,000-4,000MHz) frequencies require inductances of less than 5nH, which can be produced by less than two turns.

However, when printed d.c. bias lines are also included on the substrate, it is necessary to provide an r.f. choke (low-pass filter) to keep the r.f. out of the bias circuit, and inductances of 20-40nH are commonly used. The inductance of Fig. 7 is of this type, having an inductance of about 20nH within an external dimension of 0.110in. Square or rectangular spirals have a higher inductance per unit area than their circular counterparts, but exhibit slightly lower Q factors.

Connection to the centre of multi-turn spirals can be made either by a conductor track deposited beneath the coil and insulated from it or by a bonded wire looped directly over the turns. In this way, some circuit tuning is possible either by shorting out part of the turns or by bonding to different parts of the spiral.

Capacitors

Capacitors can be produced by the thinfilm, vacuum deposition technique in the two forms shown in Fig. 6(b) and (c). The interdigital capacitor is economical to construct as it involves no more processing steps than are normally used with microwave i.cs, that is, one plating and one etching operation. On the other hand, the parallelplate capacitor involves the deposition and etching of both a dielectric layer and another conductor. The interdigital capacitance is produced by the r.f. fringing fields across the gap between fingers and means that, in order to achieve usable values, the spacing must be small. Typically, the finger gaps range from 0.0005in to 0.001in and, by also varying the number of fingers, capacitance values of up to 1pF are conveniently produced. Having to use such small gaps means that, for consistent results, great attention has to be paid to the substrate surface finish, conductor thickness and the photo-etching process. Values of capacitance less than 1pF generally find circuit application at frequencies into X band (8,200-12,400MHz) and this type of structure can still remain a lumped element at these frequencies. Should higher values of capacitance be required with this interdigital construction, a ground plane can be introduced on the other side of the substrate. This has the effect of producing a transmission line structure like microstrip and, indeed, is particularly suitable when lumped element and distributed element circuits have to be combined. Shunt elements of capacitance are produced, giving lumped element values of several pF.

Considerably higher values of capacitance can be produced by using the parallel plate type of Fig. 6(c); it does; however, involve additional deposition and etching. For compatibility with the vacuum deposition process, a commonly used dielectric material is silicon dioxide (S_iO₂), although this has its own problems in achieving a non-porous layer having good adhesion.

The capacitance of this type of structure is directly proportional to the area and, with a good quality SiO₂ film, one micron (40

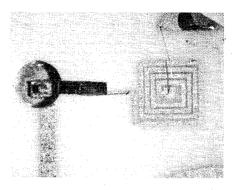


Fig. 7. For higher values of inductance, the simple ribbon can be wound into a spiral. This one is photo-etched on alumina and is about 0.1in square with 0.005in lines.

microinches) thick, $0.022\mu F/in^2$ can be obtained. Thus, even a large r.f. bypass capacitor of, say, 40pF, is only about 0.040in square.

Within the range of frequencies so far mentioned, the O factors of capacitors tend to be higher than those for inductors. In the case of the parallel plate and ground plane aided types, the losses due to the dielectric must be taken into account as well as those for the conductor. However, with good quality dielectrics, most of the loss can be attributed to the conductor material and the Q factor decreases with increasing frequency. At 1,000MHz, Qs of 1,000 or more can be achieved depending on technique, falling to 100 or less at 10,000MHz. The loss of these lumped elements compares well with the loss achieved in practice with distributed microstrip circuits and this is important when designing such circuits as amplifiers. High power transistors, for example, have low output impedances of a few ohms and, besides the inherent waste of power in the external circuit, the loss limits the degree of impedance matching, and hence power output, that can be achieved.

How far lumped elements will go towards replacing their distributed counterparts and at what frequency they cease to be practicable still remains to be seen. Much more work is required, particularly in the characterisation of stray circuit reactances and the mutual interaction of many elements. At the lower microwave frequencies, though, many successful designs of active and passive circuits have been made and, here, lumped elements can afford a considerable size reduction over even the microstrip form of distributed circuit.

Further reading

Parts 2, 3 and 4 of this series have dealt with the properties of several forms of transmission line, the formation of reactive elements and microwave i.c. technology. Attention has been deliberately focused on the topical microstrip circuitry to the (unashamed) neglect of the other types. To the best of my knowledge, there is no comprehensive textbook treatment of microstrip and microwave i.cs yet available. So, for those who would like to follow up the subjects in more detail, I provide the following list of important contributions to the technical literature, together with some more general microwave texts.

General

- 1. For a first-reader on transmission line theory and components, except for stripline and microstrip: "Introduction to Microwave Practice" by P. F. Mariner, Heywood & Co. Ltd, 1961. The appendix contains an excellent introduction to vector calculus and the formation of Maxwell's equations, culminating in the derivation of the waveguide field functions.
- 2. A detailed coverage of properties of waveguide and coaxial lines with the design of reactive elements is given in: "Microwave Transmission Design Data" by T. Moreno, McGraw-Hill, 1948
- "Microwave Measurements" by H. M. Barlow and A. L. Cullen, Constable & Co., 1950. is self-explanatary and provides a good coverage of transmission line fundamentals in addition to the measurements.
- 4. A worthy tome which covers the whole of the microwave field is "Microwave Engineering" by A. F. Harvey, Academic Press, 1963. Because of its vast range of contents, each topic is only briefly sketched, but a total of over 10,000 supporting references is provided. Much of the information on microstrip has now been superseded and, at the time the book was compiled, there was not the wealth of solid state devices that there is now.
- 5. For further details on filters and directional couplers, one can do no better than: "Microwave Filters, Impedance—Matching Networks and Coupling Structures" by G. L. Matthaei, L. Young and E. M. T. Jones, McGraw-Hill, 1964. This gives the basic theory and then the practical realization of the components in microwave form with the aid of numerous worked

Stripline and microstrip

- 6. Much of the basic stripline theory is contained in a special issue of I.R.E. Trans. on Microwave Theory and Techniques, Vol. MTT3, March, 1955, and is still applicable.
- 7. The theory of coupled lines for various stripline geometries is contained in: "Shielded I.R.E.Coupled-Strip Transmission Lines" Trans. MTT3, Oct., 1955.
- 8. The impedance of microstrip is deduced in a standard article by H. A. Wheeler: "Transmission-Line Properties of Parallel Strips Separated by a Dielectric Sheet". I.E.E.E. Trans. MTT13, 1965.
- 9. Coupled microstrip line analysis is given in: "Parameters of Microstrip Transmission Lines and of Coupled Pairs of Microstrip Lines" by T. G. Bryant and J. A. Weiss, I.E.E.E. Trans. MTT16, Dec., 1968. This is suitable only for computer evaluation and a programme is now available from the I.E.E. Programme Library. 10. Further information on the properties of microstrip is contained in: "Microstrip Lines for Microwave Integrated Circuits" by M. V. Schneider, Bell System Tech. J., May-June
- 1969 and 11. "Losses in Microstrip" by R. A. Pucel, D. J. Massé and C. P. Hartwig, I.E.E.E. Trans. MTT16, June 1968.

Lumped elements

12. "Status of Lumped Elements in Microwave Integrated Circuits" by M. Caulton, B. Hershenov, S. P. Knight and R. E. DeBrecht, *I.E.E.E. Trans.* MTT19, June 1971.

13. "Lumped-Circuit Elements at Microwave Frequencies" by C. S. Aitchison *et al. I.E.E.E.* Trans. MTT19, Dec. 1971.

14. "Effect of Process Variables on the Microwave Conductivity of Thick-Film Conductors by J. R. Corkhill and E. R. O. Donnell. Electronics Letters, 5th Oct. 1972. 15. "Electrical and Electronic Engineering" by

F. E. Terman, McGraw-Hill, 1943.

World of Amateur Radio

Aerial gain — fact or fiction?

Many v.h.f. operators use aerial arrays that are designed to give 10 dB or more gain relative to a half-wave dipole. But sometimes the gain is more theoretical than real — many factors can in practice reduce the gain achieved. Should more effort be made to check the gain figures? For several years American rallies and conventions have featured aerial gain contests with amateurs bringing along their arrays to compete under specified conditions. Winning entries may feature gains as high as 20 dB or more.

Such contests are not often held in the U.K. but we were interested to read in Break-In (journal of the New Zealand association of radio amateurs) some detailed results of a 144 MHz aerial measuring contest held recently at Christchurch — results which we suspect must have come as a shock to some entrants. For example a nine-element quad beam had a 30 dB front-to-back ratio, but its forward gain over the reference dipole was only 3 dB; a 14 dB f/b ratio on a 7-element Yagi array was accompanied by only 1 dB forward gain, whereas a 6-element Yagi with 10 dB f/b ratio had an 8 dB gain. A four-element Yagi with an f/b ratio of 4 dB actually had a loss of 11 dB compared with the dipole.

These measurements were made using a low-power transistor transmitter (100 mW output with 40 dB attenuation) spaced a quarter of a mile from the receiving site with both aerials on 20 ft aluminium masts and using a high-grade variable attenuator. Fixed attenuators were used either side of the variable attenuator to help overcome the effects of mis-matches, and the power of the transmitter was kept low to prevent direct leakage into the receiver.

Licences and operational news

Two relaxations have been made by MPT in the terms of the Class C licences issued to foreign amateurs for operation in the U.K. In future, these operators will be able to use r.t.t.y. (radio teleprinting) and to operate on telephony in any recognised

spoken language instead of being confined to the English language.

Class B amateur licences for v.h.f. operation now number over 3850 and represent more than 20 per cent of the total of about 18,650 U.K. licences for fixed-station operation. The number of new Class B licences continues to increase considerably faster than the general Class A licence.

Since the re-siting of the 144 MHz GB3PI repeater station — licensed experimentally by the MPT for one year on a 200 ft tower one mile west of Barkway in north Hertfordshire the operational range has increased considerably. Mobiles at distances up to 45 miles (60 miles to the north) are now working through GB3PI, so allowing contacts to be made between vehicles as far apart as Lincolnshire and north London. The present experimental licence expires on August 11 and a decision will have to be made soon whether the experiment will continue and perhaps be extended to include other repeaters and u.h.f. facilities, or whether the experiment be wound up.

The Oscar 6 satellite repeater schedule has again been changed; the transposer is now usually working only on Thursdays, Saturdays and Mondays. Two more British amateurs have recently obtained the Oscar 6 "1000" Award: D. E. Davies, GW3FSP of Bridgend and R. G. D. Holmes, G6RH of Bexley, Kent.

R.S.G.B. Diamond Jubilee

The London Wireless Club - later to become the Wireless Society of London and then (in 1922) the Radio Society of Great Britain — was formed on the evening of July 5, 1913 at a meeting of about half-a-dozen amateurs at the home of Rene Klein (RKX later G8NK) at 18 Crediton Road, West Hampstead. In its August 1913 issue, Wireless World commented: 'Londoń has hitherto been without a wireless club, but at a meeting held on July 5 an association was formed under the title of "The London Wireless Club" having for its object the "bringing together of all amateurs interested in wireless telegraphy and telephony".'

In Brief

Derby and district amateur radio society recently ran a direction finding contest in which participants worked on foot rather than the usual combined "car/foot" hunt. This was done to give younger members a chance to take part on their own; the society has also co-opted to its committee a junior member, Robert Bebe. . . . Another broadcast intruder in the 3.5 MHz band is HCVG8, operating from the Galapagos Islands on 3520 kHz. . . . The BBC World Radio Club, which recently enrolled its 14,000th member, is now transmitted on medium-waves (1088 kHz) at 2315 G.M.T. on Fridays instead of 2345 G.M.T. After $2\frac{1}{2}$ years as producer Joy Boatman has handed the programme over to Reg Kennedy, with Colin Marchant and Henry Hatch, G2CBB, continuing as regular participants. . . . A group of Swedish amateurs (SM7AED. SM7DBI and SM7FJE) will be operating at weekends, including early mornings, this summer from a site 215 metres a.s.l. with high-power c.w. (144.011 MHz) and s.s.b./c.w. (145.375 MHz). . . The R.S.G.B. national mobile rally takes place at Woburn Abbey on Sunday, August 5. The rally will include a large trade exhibition; R.S.G.B., B.A.R.T.G. and "Raynet" information stands; a GB3PI repeater demonstration stand and a "bring-and-buy" stand, all in a large marquee. Talk-in stations will include GB3RS (1910 kHz a.m., 3700 kHz s.s.b.); G3VHF (70.26 MHz a.m., 144.48 MHz f.m., 145 MHz a.m. and 145.41 MHz s.s.b.); and via GB3PI. . . . July mobile rallies include: South Shields (July 1); Newquay, Cornwall and Upton-on-Severn (July 8); Scarborough (July 15); and Ipswich (July 22). . . Two U.K. microwave records were broken when Les Sharrock (G3BNL) and A. Wakeman (G3EEZ) made contacts over the 98-mile path between Dartmoor and the Prescelly mountains, South Wales, on the 13, 9, 6 and 3 cm bands. Excellent signals were received on all bands. . . . Twelve v.h.f. repeaters are in operation in Australia: New South Wales (5), Victoria (3), Queensland, South Australia, Western Australia and Tasmania with input and output channels separated by 500 kHz.... Polish stations on the 432 MHz band include SP2DX, SP2AOZ, SP3BBN, SP6LB and SP6BPR. . . . The June 30 eclipse propagation experiment (see last month) will be based on transmissions from 5T5SOL operating as a beacon station on 14050 and 21050 kHz from 0930 to 1130 G.M.T. . . . The MPT says there is a growing tendency for stations operating in s.s.b. nets to ignore the requirement that callsigns should be announced at least every 15 minutes. . . . The Barking Radio Electronic Society will operate an amateur station at the Dagenham Town Show on July 7 and 8 on all bands from 1.8 to 145 MHz, and with GB2DTS as the talk-in station on 145

Industrial Electronics — 4

Electronic weighing

by Richard Graham

It is a remarkable thought that the kind of measurement which must have been one of the first that man ever made is the one which has resisted the attentions of electronic engineers the longest. It is probably not much more important to weigh things accurately today than it was when the brontosaurus steaks had to be shared out equally, but it is now considered to be insufficient to read the weight from a weighing machine; the result of the measurement must be printed, transmitted, used in computation or used to control a blending process.

The call for electronic weighing is therefore not primarily motivated by its potential accuracy, but rather by its direct provision of an electrical signal — analogue or digital — which can be used to fulfil the ancillary functions of weighing. This is, perhaps, just as well, because the achievement of accuracy in an electronic weighing system is a difficult and expensive business; only in the last few years have electronic machines reached a performance comparable in accuracy to that of the mechanical type of scale.

Hybrid weighers

The first attempts at deriving an electrical output from a weighing process were, as seen in Fig.1, hybrid in nature. The spindle of an ordinary dial scale was fitted with a rotary displacement transducer, which produced either analogue or digital signals corresponding to the position of the scale pointer. The transducer could be either a photocell-scanned diffraction grating or a refined type of slide-wire potentiometer, the former possessing the considerable advantage of friction-free operation and direct digital output.

These systems combine the advantages (and the disadvantages) of electrical and mechanical operation, and the feeling was that the benefits of electronics were not being properly exploited. For instance, the construction of a mechanical weighbridge consists of unbelievably large lumps of steel, supported on knife-edge bearings. When new, the whole edifice is capable of weighing extremely accurately and with excellent repeatability. Unfortunately, this state of affairs does not last indefinitely; knife edges and the dial mechanism wear and the pointer begins to "stick", maintenance is therefore required. Moreover,

since the platform of the weighbridge must usually be flush with the ground, a sizeable hole in the ground is needed to house the lever mechanism.

On the other hand, a purely electronic weighbridge consists of the platform, a number of load cells, and a box of instrumentation, as shown in Fig. 2. Admittedly, the pit remains, but can be considerably shallower and, in some cases, can be dispensed with altogether, a ramp being used instead. The output of the instrumentation can be either analogue or digital, and no "sticking" will be observable.

Electronic weighers

The device which has made possible the present state of electronic weighing is the strain gauge used in a load cell. Other types of load cell have been proposed, but strain-gauge cells presently reign supreme.

In essence, the device is simple, but it has taken at least twenty years to reach the current state of parity with mechanical weighers, and even now there are problems to be solved. After many years of development, the strain gauge has now stabilized into several forms. It is yet another illustration of a simple idea which is ferociously difficult to implement, and its development was retarded by the fact that most of the early workers in the field spent a lot of time industriously measuring and recording the wrong parameters, neglecting vital information.

If one considers a piece of metallic material of length L, cross-sectional area A and specific resistance p, its resistance end-to-end is expressed as pL/A. If it is deformed, becoming shorter and fatter, its resistance therefore decreases, deformation by stretching giving an increase in resistance. To apply this effect to a weigh-

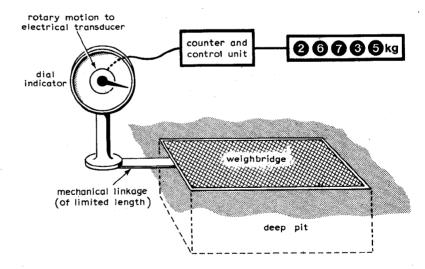


Fig. 1. A hybrid mechanical/electrical weigher.

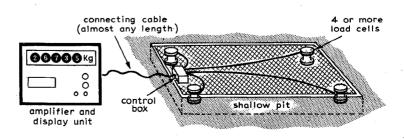


Fig. 2. Electronic weighing system.

ing system, the strain gauge is used in conjunction with a deforming spring structure, the load cell. Fig. 3 shows one form of load cell, the "active" part being the narrow centre section. As a load is applied to the button at the top, the neck becomes very slightly deformed or squashed — by a matter of a few hundredths of a centimetre decrease in height. This deformation is accurately proportional to the applied load until the elastic limit is reached, and is measured by the strain gauges bonded to the neck. The gauges, in the simplest configuration, are arranged in the form of a wheatstone bridge and the output signal is around 2mV per input volt. The signal is amplified, usually digitized and displayed. Many companies use a dial display, but this always seems to be a throwback in much the same way as cars having their engines at the front because that is where the horses used to

That is a typical electronic weigher in principle. As may be suspected, there is a lot more to it than that. For instance, a load cell would, if precautions were not taken, do a reasonably good job of being a thermometer; it could also suffer from the bonding of the gauges being imperfectly done and the ingress of moisture would not help. The metal of the billet must be of the correct type and the proper application of the load to the cell is a matter of vital importance.

For normal, mid-range temperature operation, the foil type of gauge is used. Fig. 4 shows the general arrangement of such a gauge. The gauges are manufactured by a process similar to that by which printed circuits are made. An extremely thin (about 5 microns) strip of constantan, or whatever metal is chosen, is coated on one side with synthetic resin and on the other side with a photo-sensitive resist. The resist is exposed to the required pattern through a negative, the exposed parts hardened and the rest removed. The remaining pattern can be etched away chemically, leaving the final gauge on a synthetic resin carrier, which is then bonded to the load cell billet and leads are welded or soldered to the pads. Resistance wire gauges are used for higher temperatures and many different patterns of gauge are used for other applications.

Two-component bonding materials are often used, consisting of a monomer and hardener, polymerization taking place on mixing and setting occurring in a short time with heat applied. The important features of a bonding agent are its life and creep properties, sensitivity to moisture and resistance to radiation. (Creep is the effect observed when a suddenly-applied load is reflected in a gradual approach to the final resistance value of the gauge.)

It is obvious that at least three load cells must be used to form a stable support for the weighing platform, and it is usual to use a minimum of four and sometimes eight, depending on the size of the platform. The way the cells are connected together to give a signal output determines the sensitivity of the system, and also, to some extent, the type of supply used.

If cell outputs are connected in series,

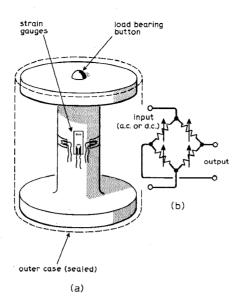


Fig. 3. A typical load cell.

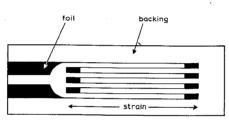


Fig. 4. One type of etched-foil strain gauge.

with separate drive inputs, then the combined output is that of one cell. The output impedance is high and three supplies are required. On the other hand, if the outputs are in parallel, the signal is 1/n that of one cell, where n is the number of cells. The output impedance is low and only one supply is needed. It is common practice to use a series-parallel combination of cells, giving reasonably low impedance, adequate sensitivity and not too much trouble with power supplies.

The type of cell described is one of several different styles: they are also made in the form of rings and horizontal beams, depending on the range of weights to be measured and on the method of applying the compressive or tensile force to the transducer. Accuracies are usually grouped into two classes generalpurpose and precision. A precision cell, used in weighbridges which must satisfy official requirements, is capable of calibration error, non-linearity, hysteresis, and creep of around 0.05% of full output, while the general-purpose variety will give calibration errors and non-linearity of 0.1 to 0.25% full rated output. Temperature effects vary from 0.001% to 0.005% of rated output.

That, then, is a very brief description of the basic weighing element. There remains its application to the business of weighing, the ancillary instrumentation and the use of complete systems in industry. I will try to cover as much of this as possible in the next article.

Announcements

A course for experienced standards engineers will be held at the University of Edinburgh from 25th to 27th June under the joint sponsorship of the Standards Associates Section of the British Standards Institution, 2 Park Street, London W1A 2BS and the University's Centre for Consultancy and Liaison. It is designed to bring standards engineers up to date on standardization matters and provide a forum in which they can discuss their problems, and exchange ideas.

Burndept Electronics (ER) Ltd, St. Fidelis Road, Erith, Kent DA8 1AU, has acquired the land mobile radio communications business of Cossor Electronics. The equipment involved comprises the "Consort" solid-state control system and u.h.f. fixed point-to-point links for communication and telemetry.

Sales office of ITT Components Group's Thermistor Division, formerly in Harlow, has moved to Steven Street, Taunton, Somerset. It is now at the same location as the factory.

Full qualification approval to the Post Office D3000 Class One specification has been gained by the Electronic Components Division of Ferranti Ltd, Gem Mill, Chadderton, Oldham, Lancs, for their low-power, t.t.l. integrated circuits. To obtain this approval, it must be demonstrated that cumulative failures over 40 years will not exceed 2% in an environment of 70% relative humidity, and 55°C ambient temperature.

Two new undersea telephone cables are to link Britain and France in a project which will cost between £5M and £6M. The British Post Office and the French Ministry of Posts and Telecommunications are to share equally the cost of the new cables, which will double the number of telephone circuits between the U.K. and France.

Thorn Colour Tubes Ltd, Skelmersdale, Lancs, have announced a £4.5M expansion programme. Current output is running at 750,000 tubes per year and the new expansion programme, designed to take effect early in 1974, will eventually increase output to 1.2M tubes per year.

British Aircraft Corporation, Electronic and Space Systems, Brooklands Road, Weybridge, Surrey KT13 0RN, have received a £1.2M export order from Hughes Aircraft Company for the manufacture of satellite subsystems for the latest global communication satellites Intelsat IVA.

Two new Boyd Line trawlers being built in Poland, B420/1 and B420/2, are both to be supplied by Marconi Marine, a GEC-Marconi Electronics company, Marconi House, Chelmsford CM1 1PL, with major electronics packages which will include communications equipment and navigational aids.

ITT Semiconductors have appointed W. S. McMillan & Co, 61 Colvilles Place, Kelvin Industrial Area, East Kilbride G75 OTA, Glasgow, as franchized distributor for Scotland and the N.E. of England. Also appointed as specialist distributor for the radio and TV industry are Best Electronics (Slough) Ltd, Michaelmas House, Salt Hill, Bath Road, Slough, Bucks SL1 3UZ.

Marconi Communication Systems Ltd, a GEC-Marconi Electronics company, Marconi House, Chelmsford, have received a £3½M contract from the Yugoslav Joint Radio Television Committee, to provide the largest single share of an order for studio and transmitting equipment to put Yugoslavia's second TV service on the air.

Plessey Semiconductors has appointed Gothic Electronic Components, a division of ERD (Midlands) Ltd, Beacon House, Hampton Street, Birmingham 19, as authorised U.K. distributors for the complete range of Plessey integrated circuits.

BURNDEPT HAVE PLANS TO SPEED UP A BUSINESS ON THE MOVE



Burndept's new Personal Radio-telephone is the most versatile unit yet designed, It's small and light, and has a wide range of accessories. So it can be used by anybody anywhere—airlines, construction firms, the oil industry—all find a use for it. The Burndept mobile unit is easily installed in lorries, cars, fork lift trucks. Burndept Radio-telephones are available in the VHF frequency bands (either FM or AM) and in the UHF band for greater penetration and clarity in urban areas. Burndept back up their systems planning with a most efficient after-sales service.

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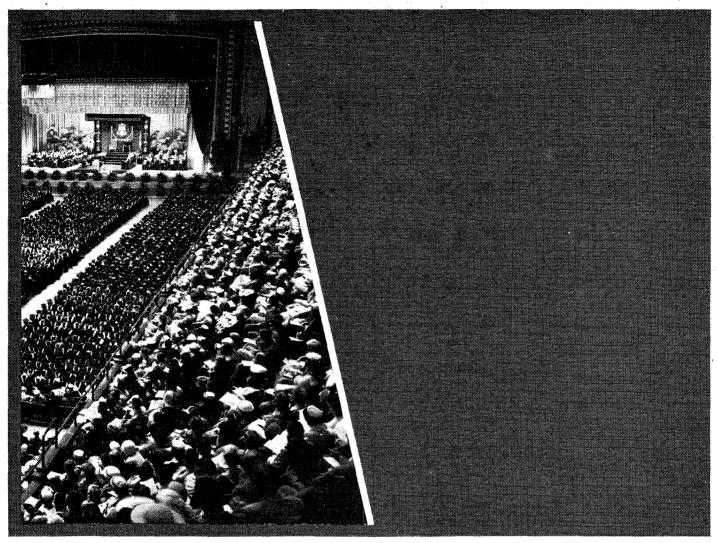




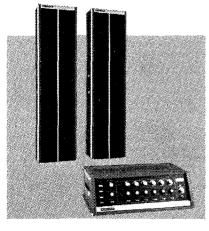
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Experiments with operational amplifiers

11. An op-amp used as a phase sensitive detector

by G. B. Clayton* B.Sc., F.Inst.P.

A phase sensitive detector (p.s.d.) is a system which produces a d.c. output signal in response to an a.c. input signal of a defined frequency equal to the frequency of an a.c. reference signal. The d.c. output is proportional to both the amplitude of the a.c. input signal and the cosine of its phase angle relative to that of the reference signal. P.s.ds are used as synchronous rectifiers in chopper d.c. amplifiers; they are also used for the accurate measurement of small a.c. signals obscured by noise. ¹

A p.s.d. acts essentially as a multiplier giving an output proportional to the product of the input signal and the reference signal. It is instructive to analyse this multiplication process. We consider a sinusoidal signal $e_s = E_s \sin{(\omega_s t + \phi)}$ multiplied by a square wave reference signal. The square wave is assumed to be symmetrical about zero and to have unit amplitude and unity mark-space ratio.

The Fourier series for a unit amplitude square wave is

$$v_r = \frac{4}{\pi} \left[\sin \omega_r t + \frac{1}{3} \sin 3\omega_r t + \frac{1}{5} \sin 5\omega_r t + \dots \right]$$
(11.1)

and

$$e_s v_r = \frac{4}{\pi} E_s \left[\sin (\omega_s t + \phi) \sin \omega_r t + \sin (\omega_s t + \phi) \frac{1}{3} \sin 3\omega_r t + \dots \right]$$

$$= \frac{2}{\pi} E_s \left[\cos \left[(\omega_s - \omega_r) t + \phi \right] - \cos \left[(\omega_s + \omega_r) t + \phi \right] + \frac{1}{3} \cos \left[(\omega_s - 3\omega_r) t + \phi \right] - \frac{1}{3} \cos \left[(\omega_s + 3\omega_r) t + \phi \right] + \frac{1}{5} \dots \text{ etc.} \right]$$

$$(11.2)$$

It can be seen that the product gives rise to d.c. terms for signal frequencies $\omega_s = \omega_r$; $3\omega_r$; $5\omega_r$; etc.

For
$$\omega_s = \omega_r$$
 the d.c. term is $\frac{2}{\pi} E_s \cos \phi$

In a p.s.d. the multiplication process is followed by low pass filtering. The low pass filtering may be performed by a simple RC filter which is used to attenuate all a.c. components of the product. The effective bandwidth of the system is then set at 2/T radians/sec. where T = CR, the time constant of the filter.

The noise rejection properties inherent in phase sensitive detection arise in large part from the extremely narrow bandwidth which is so readily obtained by selection of filter time constant. A p.s.d. acts as a rectifier tuned to the reference frequency, but its noise rejection properties are in fact superior to a filter rectifier combination. It would be quite impractical to obtain the narrow bandwidth of a p.s.d. with a conventional filter.

Noise passed by a conventional filter circuit gives rise to d.c. in the normal rectification process. In a p.s.d. unwanted signals, like random noise, produce fluctuations about the d.c. output level given by the wanted signal at the reference frequency.

Many p.s.d. designs of varying degrees of complexity have been described in the literature (Refs. 2, 3, 4, 5). In general it is convenient if the signal input, the reference voltage input and the output from the detector share a common earth line. This condition is not fulfilled in all p.s.d. designs but

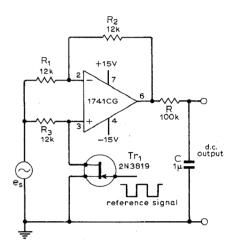


Fig. 11.1 Op-amp phase sensitive detector.

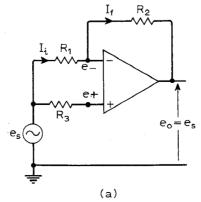
may be readily fulfilled by using an operational amplifier as a p.s.d.

In the circuit of Fig. 11.1 the closed loop gain of the operational amplifier is switched between plus and minus unity by a square wave reference signal applied to the gate of the f.e.t. Tr_1 . The circuit is very convenient for illustrating p.s.d. operation.

Conditions in the circuit with the f.e.t. biased off and on are shown in Figs 11.2(a) and 11.2(b) respectively. Referring to Fig. 11.2(a) and assuming the amplifier behaves ideally we have

$$e = e = e$$

Thus $I_i = I_f = 0$ and e_o must be equal to e_s . The closed loop gain is plus unity and



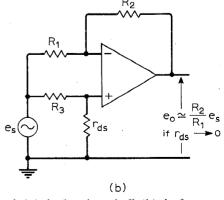


Fig. 11.2 Conditions in the p.s.d. of Fig. 11.1 with (a) the f.e.t. biased off, (b) the f.e.t. biased on.

for $\omega_s = 3\omega_r$, the d.c. term is $\frac{2}{\pi} E_s \frac{1}{3} \cos \phi$... etc.

^{*}Department of Physics, Liverpool Polytechnic.

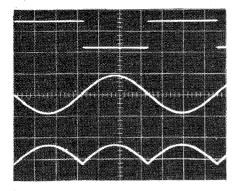


Fig. 11.3 Signal (sine wave) and reference (square wave) in phase. (Upper trace 10V/div.; middle and lower traces 5V/div.; horizontal scale 2ms/div.)

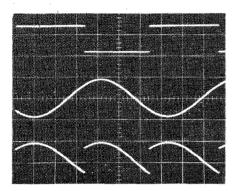


Fig. 11.4 Signal (sine wave) and reference (square wave) phase shifted by 45°. (Scales as in Fig. 11.3.)

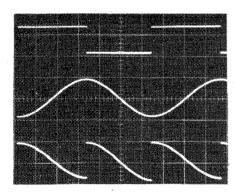


Fig. 11.5 Signal (sine wave) and reference (square wave) phase shifted by 90°. (Scales as in Fig. 11.3.)

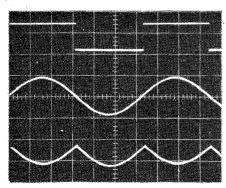


Fig. 11.6 Signal (sine wave) and reference (square wave) phase shifted by 180°. (Scales as in Fig. 11.3.)

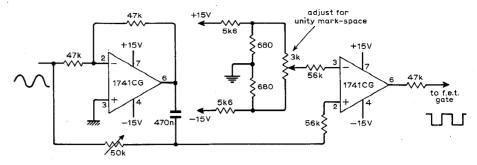


Fig. 11.7 Phase shifting and squaring circuit

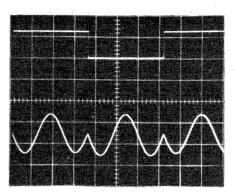


Fig. 11.8 Response of p.s.d. to 3rd harmonic of reference signal.

under the conditions in Fig. 11.2(a) its value is independent of resistor values.

In Fig. 11.2(b) r_{ds} represents the effective drain source resistance of the conducting transistor. If $r_{ds} \rightarrow 0$ the non-phase-inverting input terminal is effectively earthed and the closed loop gain is equal to $-R_2/R_1$. A nonzero value for r_{ds} allows a small fraction of the input signal to appear at the non-phase-inverting input terminal of the amplifier and R_2 must be made slightly greater than R_1 for a closed loop gain of minus unity.

The action of the p.s.d. is illustrated by the oscillograms, Figs 11.3 to 11.6, which show circuit waveforms for various phase relationships between input and reference signal. In each case the upper trace shows the reference square wave which is applied to the gate of the f.e.t., the middle trace shows the sinusoidal input signal and the lower trace shows the waveform which appears at the output terminal of the amplifier. The average value of the output is read by a d.c. voltmeter which follows the low pass filter *RC*.

Note that when the square wave is at its negative value the f.e.t. is switched off and the sinusoidal signal is multiplied by plus unity. When the square wave is at its upper level the sinusoidal signal is multiplied by minus unity. The unity amplitude square wave reference multiplier of eq. 11.1 is phase shifted by 180° on the actual reference square wave.

In order to check the relationship

$$E_{d.c.} = \frac{2}{\pi} E_s \cos \phi$$

it is convenient to derive the phase shifted reference square wave and the input signal from the same signal source.

A phase shifting and squaring circuit suit-

able for this purpose is illustrated in Fig. 11.7.

It is instructive to observe the behaviour of the p.s.d. for input signals at frequencies other than that of the reference signal. Two separate sources are required for this purpose, a reference square wave and an input sinusoid. The oscillogram in Fig. 11.8 shows waveforms which illustrate the response of the system to an input signal at a frequency equal to the third harmonic of the reference signal. It is clear that the output waveform has a non-zero average. Its value is

$$\frac{2}{\pi}E_{s}\frac{1}{3}\cos\phi$$

where in this case $\phi = 180^{\circ}$.

References

- 1. Danby, P. C. G., Electronic Engineering, January 1970.
- 2. Williams, P., J. Sci. Instrum., 1965, 42, 474-476.
- 3. Lacy, J. G., Electronic Engineering, March 1967.
- 4. Danby, P. C. G., Electronic Engineering, December 1968.
- 5. Linsley Hood, J. L., *Electronic Engineering*, April 1970.

About People

Professor John Bardeen, For Mem.R.S. In recognition of his development of the basic theory leading to the invention of the transistor and of his research on the theory of solids, John Bardeen has been elected a Foreign Member of the Royal Society. In 1956, Professor Bardeen shared with William Shockley and Walter Brattain the Nobel Prize for physics, and was awarded the I.E.E.E. Medal of Honour in 1971. He is Professor of Electrical Engineering and Physics in the University of Illinois, Urbana.

Richard Foxwell, C.B.E. has been appointed a director of Marconi Instruments. Mr. Foxwell, who was a founder and chairman of Wayne Kerr, was a past president of the Scientific Instrument Manufacturers Association, is deputy chairman of the Advisory Council on Calibration and Measurement, deputy chairman of the Conference of the Electronics Industry and U.K. co-chairman of the Anglo-Soviet Joint Working Group on Instruments. He has recently been nominated as the U.K. representative on the Presidents Committee on the Comité des Industries de la Mesure Electrique et Electronique de la Communauté (CIMEC).

The Metre Waves Now

A progress report on amateur-radio activity above 30MHz

by J. Hum G5UM

Officially, v.h.f. begins at 30MHz, u.h.f. at 300MHz and so on upwards. Colloquially, the whole of this spectrum is known as the metre wavelengths by the amateur radio transmitting operators who inhabit it, to distinguish it from the "h.f. bands" of 30MHz and below. We shall use the terminology in this broad sense for the

remainder of this article.

Over the last half dozen years the exploration of the metre wavelengths by U.K. transmitting amateurs bears some resemblance to the pioneer work done by their forebears almost half a century earlier in opening up the ostensibly useless short waves. But there is a significant measure of difference: in the "way back" days amateur signals were almost the only ones to be heard on the virtually vacant intercontinental wavebands. Today the world above 30MHz has a population problem so intense that official specifications for professional communications equipment become tighter almost year by year. Once-spacious 50kc/s channelling (in the days of kilocycles, before "hertzes" came in) gave way first to 25kHz and then to $12\frac{1}{2}$ kHz bandwidths that told their own story of the congestion which had built up on the very high frequencies.

To the amateur transmitting practitioner these circumstances carry certain implications. No longer alone on the metre wavelengths, as his father was on the short waves, he is well aware of the need to justify his retention of them under such pressures. But what he also observes is the professionals' quite understandable preoccupation with providing as near 100% reliable communication as the noted intransigence of electronic equipment will permit, both in terms of actual electromechanical reliability and in the sustaining of high level voice communication come what may, and no outages (reversion by the customer to landline communication is not only a stigma on the escutcheon of the business-radio manufacturer but carries the awkward possibility that it might become permanent).

This is not to say that the amateur operator puts a low priority on reliability: in a situation where large numbers of inter-city and inter-county nets on the 144 and 432MHz bands go into action nightly, instant communication is a sine qua non. But it is to say that the amateur v.h.f. communicator today detects, just as his

father did fifty years ago, a number of challenges to his technique which in the nature of things can be of only minimal interest to the professional providers of short haul traffic facilities.

Anomalous propagation

Just above the upper limit of television Band I lies the amateur 70MHz allocation, popularly known as "Four Metres", apparently capable of giving no more than line of sight range if considered in conventional engineering terms. To questing experimenters it has shown that it can give much more. Early in their experience of "Four" (it was granted to them 17 years ago), British amateurs discovered that it held propensities for anomalous propagation, even with transmitter inputs restricted to the 50W allowed on this band, that were truly surprising. The value of the band to amateur v.h.f. men, and the value of the propagation observations they were conducting on it, persuaded the licensing authority progressively to extend the available space from a bare 200kHz to 675 kHz today. Latterly, a few overseas countries have followed the U.K. lead and made this special band available to nonprofessional users, "special" because it is not an official I.T.U. amateur allocation.

The effect of these concessions was startling. Communication from the U.K. to Iceland in one direction and Gibraltar in the other became almost a matter of routine during the equinoctial months of certain years when it was possible to use the sporadic-E manifestations that generate reflective ionized patches along the propagation paths.

Another aid to long distance communication on supposedly optical frequencies is the tropospheric or warm weather effect, generally associated with stable high pressure weather systems and temperature inversions, and the appearance on the domestic television set of the "We regret continental interference" caption. Whereas tropo-propagation can be predicted by the sixth sense which amateur v.h.f. men develop in relating weather conditions to radio conditions - a look at the barometer and then the sky is enough — it is not so easy to forecast when sporadic-E will occur (except, broadly, in the equinoctial seasons). When it does it is supremely exciting and brings hundreds of v.h.f. workers off the garden - or early home from work

— in order to participate in what is always a short-period occurrence: a characteristic of sporadic-E is its habit of "turning on" without warning and later of "going out like a light".

Equally intriguing is the possibility of communication via the Auroral curtain upon its appearance after periods of intense solar activity. Radio noise from the sun, a harbinger of aurorae, alerts the amateur observer to the possibility that Northern Lights may be present even though they may be invisible. On these occasions directional aerials pointed northwards transmit v.h.f. signals to be returned by the aurora and received on other aerials orientated the same way. Such signals are rendered characteristically rough in tone quality by the discrete nature of the reflecting surface ("super multipath" propagation), and for this reason a.m. telephony becomes unintelligible. Either c.w. or single sideband telephony is essential.

All these propagation abnormalities occur in the 70MHz band and have been thoroughly exploited by amateur operators. Their appearance in "the next band up", 144MHz, is less frequent, and on 432MHz (70cm) and 1296MHz (23cm) rare indeed except under special conditions of "warm weather tropo" that extend the range of the ultra-high frequencies in a manner likely to bring delight to the amateur but dismay to the professional planners of short haul communication.

Class B licence

Of the four bands so far referred to, that extending from 144 to 146MHz is by far the most heavily populated. A prime reason for this is the advent of the Amateur Class B licence which requires no morse qualification, but confines the holder to the bands 144MHz and upwards. When introduced in 1964 this licence permitted operation on telephony in the 432MHz (70cm) band and above. In March, 1968, the then Postmaster General announced that in response to representations made by the Radio Society of Great Britain he had decided to include the 144-146MHz band within the terms of the Class B licence. An immediate acceleration in the rate of issue of Class B licenses so galvanised the activity and occupancy of "Two" that now it is no

longer necessary to tune over the band for a long time before a contact can be made: at any time of the day there will always be somebody on — and the chances are that he will be a Class B man using the characteristic G8-plus-3 callsign. This callsign block, which started at G8AAA in 1964, is approaching the end of the G8HAA-HZZ series, to bring the total of Class B licensees to almost 4000, most of them having come into amateur radio within the last four years, and now operating primarily in the 144-146MHz band.

It is not surprising in these circumstances that the 432MHz band is being rediscovered by large numbers of operators who "want a change from Two". Nor is it to be wondered at that a lively awareness exists that heavy band occupancy demands good manners allied to impeccable equipment design (whether home built or bought) if users of the v.h.f. world are to remain at peace with one another.

On the latter point the R.S.G.B recommends its members to strive for a rejection figure of -90dB in respect of unwanted transmitter products which is -30dB more stringent than professional requirements.

Global DX

Remaining a Class B licensee and staying permanently on telephony on the metre wavelengths suits a sizeable proportion of the G8-plus-3 operators. Others regard Class B status as an interim one preparatory to passing the morse test and acquiring the Class A licence that confers permission to use the lower frequency bands.

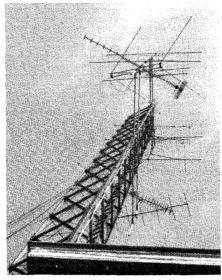
These operators are seldom lost to v.h.f. All have metre-wave equipment dating from their Class B days, and they continue to use it between spells of savouring the condition of the world below 30MHz.

To acquire a Class A licence, and the freedom to use the morse code, greatly extends the range over which a v.h.f. operator may expect to work, for telegraphy provides readability at remote points where telephony, even penetrative single sideband, can no longer be resolved.

Additionally, there are two sophisticated areas of communication ("rarefied" is probably a more appropriate adjective) where the use of telegraphy is mandatory. These are meteor scatter and Moonbounce.

Meteor showers release in the upper atmosphere tiny patches of ionization which may be used for radio communication purposes as if they were sporadic-E accumulations. Being infinitesimally smaller and more fleeting, they produce a barely audible bounced signal at a distant point; yet several U.K. stations have communicated on 144MHz as far as Russia by this means. The key (literally) to success with meteor scatter v.h.f. propagation is the ability to effect very rapid morse exchanges over a period of perhaps some hours so that bits of received informationmay be assembled to build complete messages.

The ultimate in global DX working is to be realised by bouncing a signal off the Moon as far as the Antipodes, a feat achieved on a world wide basis by no



Although a profusion of commercial designs is available to v.h.f. radio amateurs, many continue to prefer to build their own equipment. Rather fewer go so far as the Belgian enthusiast who constructed this 65ft mast himself, together with four 10-element aerials for 144MHz and four 15-element aerials for 432MHz arranged in a box formation, plus a multiplicity of other aerials for television reception.

more than a handful of dedicated experimenters. For technical reasons the 432 and 1296MHz bands are favoured for this work, for the larger the aerial dish the better the chances of success. The challenge of this kind of ultimate DX is such that, not content with achieving it by the morse key, one or two operators have Moonbounced single sideband telephony, with by no means discouraging results.

Microwaves

In a more earthbound context the increasing activity now apparent on the amateur microwave frequencies prompts the thought that centimetric parabolas will become a more established part of the amateur radio scene than they are at present. In spite of exacting aerial alignment requirements in the 13, 9 and 3cm bands, and the cut-off effect of local topography, it is to be expected that urban chimneys will assume a modestly "spaceage" look as more aerial dishes go up.

To promote further experimentation in these amateur microwave bands the R.S.G.B. offers special certificate awards to operators who communicate over certain minimum distances in the gigahertz spectrum. These are not easily won: for example users of 13cm must provide proof of two-way communication over a distance of 500km. Even at 3cm (the "ten gig" band is for a number of equipment reasons the most popular of the microwaves) the minimum distance requirement is 150km, or effectively 100 miles, achievable in certain parts of the U.K. between line of site hilltops, rarely from home.

Healthy competition stimulated in this way is a noted incentive to experimenters

to improve their equipments and to further the state of the microwave art in general.

Artificial aids

Of the methods of metre-wave propagation so far discussed, whether by morse key or microphone, or via the Moon, meteor trails, aurorae, sporadic-E or warm weather tropo, all are achieved by natural means. Now artificial aids to extend transmission ranges are coming into the picture, especially in the shape of Orbital Satellites Carrying Amateur Radio, personified as Oscar.

It was as long ago as December, 1961, that Oscar 1, launched from a site in California, prompted the thought in the minds of many v.h.f. workers that herein lay the possibility of world wide communcation on line of site paths if the project could be suitably developed. The simple 3-transistor Oscar 1 delivered no more than 100mW into a monopole aerial, transmitted the basic telemetry symbol HI, and orbited the earth every 91 minutes. It has been succeeded by a range of amateur communication satellites of increasing sophistication, the latest of them, Oscar 6, being launched at the end of 1972, upchannel centred on 145.9MHz and a down channel after transponding at 29.6MHz. Global communication has proved to be so easy to achieve via Oscar 6 that the appearance of the satellite over the radio horizon precisely at the predicted orbital times no longer excites unusual interest.

The secret of success with Oscar 6 has been to transmit in the 2m band at the critical moment when the satellite was midway between the observer and the distant station he wished to work, possibly up to 7,000 miles away, and to listen for the transponded reply on 10m before the satellite moved out of range in a matter of ten minutes or less.

Beacon signals on 10m and 70cm planned for Oscar 6 will be a feature of its successor, Oscar 7. But instead of one translation system there will be three, two of them 2m to 10m and the third 70cm to 2m. Only one will be in operation at any given time during the planned three year life-cycle of this satellite.

Repeaters

It is not only in space that amateur-band repeater services are operational: back at ground level intercommunication will be extended by the provision of hilltop talk-through repeaters to give poor-sited operators a better chance to speak to the outside world than they have ever enjoyed before. "Poor sited" includes operators of vehicle-borne equipment.

A well sited automatic repeater station enormously extends the potential range of any mobile radio communication installation operating through it. The first U.K. repeater beacon, designed by radio amateurs at the Pye works at Cambridge, licensed as GB3PI by the Ministry of Posts and Telegraphs through the R.S.G.B., and installed at an elevated site in north Hertfordshire, has virtually quadrupled the service area to be expected by a mobile operator using the 2m band.

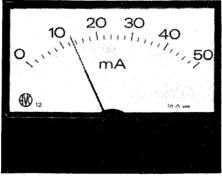
One word says our new Panel Meters are among the World's best

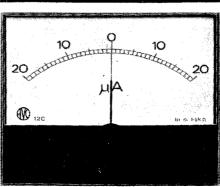
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WW-082 FOR FURTHER DETAILS

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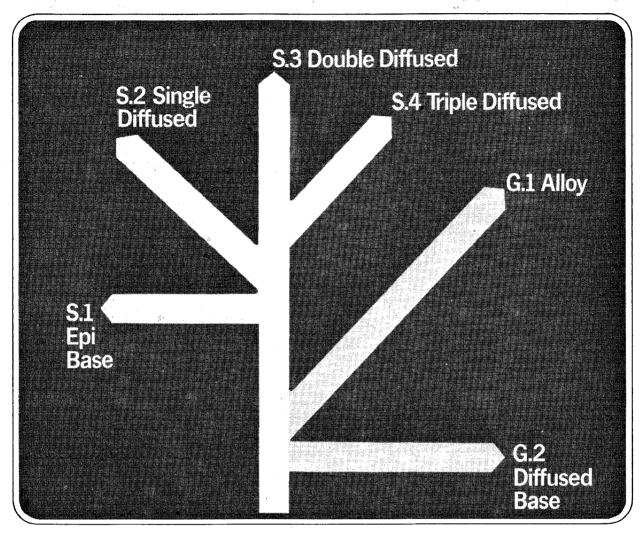
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New Products

Integrated circuit power supplies

Best Electronics announce the Westinghouse VR1 series of rectifier/regulator modules, which are fully protected against over-current and short circuits. The five standard units provide from 1.25V to 30V regulated d.c., up to 2A direct from a low voltage a.c. input, provided by the HRT-1 transformer. This transformer provides secondary taps to cover all Hyreg input voltages within 3.5%. Precise output voltages and current limiting are set by three external resistors.

For a complete power supply a Hyreg VR1 module, the HRT-1 transformer, three resistors and a smoothing capacitor are required. Prices, including post and packing of the VR1 regulator modules and the HRT-1 transformers are as follows:—VR1/1.25 — 6.5V £6.25 each, VR1/18—24.5V £6.95 each, VR1/6—12.5V £6.25 each, VR1/24—30V £6.95 each, VR1/12—18.5V £6.95 each, HRT—1 £3.95 each. Best Electronics (Slough) Ltd, Michealmas House, Salt Hill, Bath Road, Slough, Bucks SL1 3UZ.

WW 310 for further information.

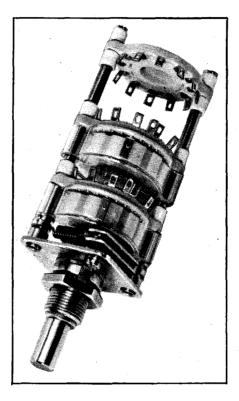
Rotary stud switch

Elma claim improvements to their 03 miniature rotary stud switch which have resulted in reducing 'switching noise' still futher and enhancing operating life — better than 25,000 rotations — to give even greater reliability. Radiatron Components, who market Elma products in the U.K., are now introducing the improved 03 version to their customers. Radiatron stress that it is mechanically interchangeable with the former type.

A choice of contact materials is offered to meet specific applications: gold flash on silver, 3μ m hard gold and 5μ m hard gold. The latter type has recently received Post Office approval.

Other significant 03 features include: Contact resistance typically $5m\Omega$, up to 24 ways on 1in wafer, up to 12 wafers, shorting or non-shorting, concentric spindles.

Design improvements have been made to the rotor and contact housing, contact



wiper, contact spring and stud, and the wafer rotor guide. Radiatron Components Ltd, 76 Crown Rd., Twickenham, Middlesex.

 $WW\,314$ for further information.

U.H.F. passive equalizer

A passive channel equaliser for four channel u.h.f. operation designed by Wolsey Electronics allows compensation for variations between the input levels of channels and losses in the distribution circuitry, so that the signal inputs to a head end amplifier system are suitably balanced. The filter system has a constant impedance potentiometer to control each channel.

Good inter-channel selectivity is claimed for this method, with excellent response curves and input and output v.s.w.r. Although the through loss is acceptable at the head end in most cases, the Wolsey u.h.f channel equalizer is line-power bypassed to allow mast head or prior amplification powered through the unit. The equalizer can be used to provide a

chosen equalizing slope to compensate for differential losses throughout a master aerial TV system, and wherever a channel filter is necessary to improve selectivity. The bandwidth of the equalizer is 9MHz at 3dB. Wolsey Electronics, Cymmer Road, Porth, Rhondda, Glamorgan CF39 9BT.

Low-pass filters

WW 317 for further information.

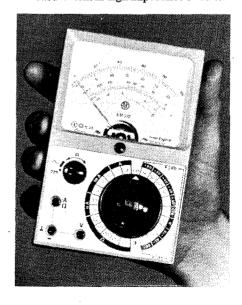
A series of low-pass filters (the FLC series) have been introduced by American Electronic Laboratories. Suited for elimination of undesired harmonics from a signal source, the FLC 1000 sector of the series is an eleven element filter with a 2MHz to 300MHz cutoff frequency range and a Chebychev response in the passband.

The FLC 3000 group of this series is a nine branch elliptic function filter with a Chebychev response in both the passband and stopband. Featuring a 2MHz to 150MHz cutoff frequency range, their greater selectivity can be used to facilitate receiver operation in the presence of a nearby transmitter. Filters in both the FLC 1000 and 3000 series have an input and output impedance of 50Ω. Low ripple design ensures a v.s.w.r. of 1.5:1 maximum for frequencies below 80% of cutoff frequency. American Electronic Laboratories, Inc., MS/1123, P.O. Box 552, Lansdale, Pa.19446, U.S.A.

WW 301 for further information.

Electronic multimeter

Avo claim to have produced one of the cheapest electronic multimeters ever for the professional market. This latest addition to the new Avometer generation is called the EM 272. A high input impedance of $316k\Omega/V$, a wide range of measurement and a good frequency response to 20kHz are offered with a.c. and d.c. voltage ranges from 30mV to 1000V f.s.d., a.c. and d.c. current from 3μ A to 3A f.s.d. and resistance between 1Ω and $40M\Omega$, the high impedance of the instrument ensuring accuracy of measurement in high impedance circuits.



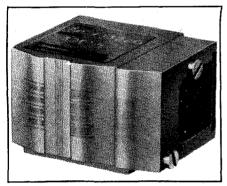
Printed circuit shunts and thick-film resistor modules provide high reliability comparable to that of the small centre-pole movements. The moulded shatterproof case, identical to that of the already popular Avometer model 72, can be held in one hand. Available from wholesale distributors at the U.K. trade price of £24.85 plus VAT £2.49. Avo Ltd, Avocet House, Archcliffe Road, Dover, Kent.

WW 316 for further information.

Cartridge/cassette recording head

To meet the rapidly increasing use of cartridge/cassette tape drives in mini computer systems, Gresham Infomag Limited have announced a new four-track read after write magnetic recording head for $\frac{1}{2}$ in tape.

Utilizing 0.064in track centres, the write head track width is 0.050in and the read width 0.030in. Bit densities up to 1600



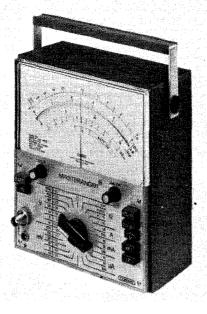
b.p.i., phase encoded, can be handled by the head, the standard product in the range being aimed at 30 i.p.s. drives. Excellent feedthrough (crossfed) and resolution are claimed for the head, whose overall mechanical dimensions are 0.650in wide \times 0.775in high \times 0.725in deep. Gresham Infomag Ltd, Weybridge Trading Estate, Weybridge, Surrey KT15 2RN.

WW 319 for further information

Multitestset

A new high input impedance multitestset has been released by Conway Electronic Enterprises. This instrument, now internationally offered, has been given the trademark registered name "Masteranger" model 639. It possesses a floating input entirely obviating user hazard particularly when measuring from circuits which are not necessarily at ground potential on the low side.

The instrument features an input resistance of $100 \mathrm{M}\Omega$ and incorporates ranging giving measurement capability from $50\mu\mathrm{V}$ up to $1500\mathrm{V}$ a.c. and d.c., plus r.f. measurement. A.c. and d.c. current can be measured from $150\mathrm{nA}$ full scale up to $150\mathrm{A}$. Resistance measurement is up to $10,000\mathrm{M}\Omega$ and dBm measurement is from $-80\mathrm{dBm}$ to $+66\mathrm{dBm}$. It is energized by



its own internal long-life batteries which can be zinc carbon, mercury, alkaline or, optionally, can be a.c. mains powered by a "line power cassette" insertable in the rear of the instrument in lieu of the battery cassette.

"Masteranger" model 639 has a six inch mirror scale and accuracy is 1.5% of full scale. The line power cassette makes use of a new type of transformer with separated windings and therefore completely isolates from each other which has resulted in implicit safety in measurement and guarantees against hazard. "Masteranger" with its total of 80 main ranges plus auxiliary ranges, will operate from batteries or $115/230V \pm 15\%$, 50/400Hz +10% and thus is useful for operation anywhere in the world and even from aircraft 400Hz supplies. The instrument is lightweight and comes complete in a genuine bridle leather carry case with shoulder and neck strap and can be used suspended from the neck of the operator ensuring hands-free operation. Conway Electronic Enterprises Ltd, 88-90 Arrow Rd, Weston, Ontario, Canada M9M 21.8.

WW 323 for further information.

Half-tone storage c.r.t.

A new cathode ray tube announced by Mullard is designed for use in laboratory oscilloscopes that require a half-tone storage facility. Designated type L14-110GH/55, it is a 14cm rectangular tube with a storage time of at least 1.5 minutes; the phosphor used to present signals without storage has a persistence of short-to-medium duration.

The new tube is only a few millimetres longer than types D14-120, D14-121 and D14-160, other dimensions being the same. Hence, the L14-110GH/55 can easily replace them in existing oscilloscopes which will then gain the facility of image storing. As the new tube has greater horizontal deflection sensitivity than the earlier types,

an improved performance is also gained in the oscilloscopes.

The L14-110GH/55 has a useful scan area of 90×72 mm (10×8 divisions of 9mm). Correction coils supplied with the tube enable the raster to be aligned with the internal graticule. Mullard Ltd, Mullard House, Torrington Place, London WC1E 7HD.

WW 318 for further information.

Modulation meter

An automatic modulation meter with a frequency range sufficient to cover all the mobile radio bands is announced by Sayrosa Engineers Ltd. The instrument, model 251, employs circuitry which eliminates manual tuning and level setting so that as long as the input signal is above a minimum level (typically 10mV), the circuitry will lock-on and provide an immediate measurement of the modulation present. F.M. deviation ranges from 3kHz to 100kHz full scale are provided and for a.m. there are full scales of 30% and 100%. Mode and range selection can be made manually, or by logic input an analogue output proportional to the measurement is provided for systems use. The unit is halfrack width, 4in height and weighs less than 8lbs. Sayrosa Engineers Ltd, P.A. Box 50, Bracknell, Berkshire RG12 4YT.

WW 307 for further information

Frequency analyser

B & K Laboratories have announced a new frequency analyser. This model, known as the 2120, is a constant percentage bandwidth analyser for use in the frequency range 2Hz to 20kHz with provision for the connection of external filters extending the range to 180kHz. It consists of a measuring amplifier combined with a built-in active filter complex which is continuously variable and can be switched into four basic modes:

- 1. As a constant relative bandwidth filter having four selectable bandwidths of 1%, 3%, 10% and 1/3 octave
- 2. As a tunable bandstop filter which will suppress any chosen frequency more than 60dB; suppression at $0.5f_0$ and $2f_0$ being less than 1dB.
- 3. As a tunable high-pass filter.
- 4. As a tunable low-pass filter.

The internationally standardized weighting networks for sound measurements — A, B, C and the proposed D network — are included and are selectable as required.

The linear frequency range of the instrument is 2Hz to 200kHz, and the measuring range from $10\mu V$ to 300V r.m.s. or 700V peak. Noise level can be reduced to $0.4\mu V$.

The many refinements incorporated in the model 2120 combine to make it a very versatile instrument indeed.

The meter and its rectifying system for instance, provides a choice of 6 indications in either logarithmic or linear form. These

are r.m.s., impulse hold, +ve peak, -ve peak or maximum peak and in addition a switch provides a choice of averaging times from 0.1 to 300 seconds as well as "fast" and "slow" in accordance with I.E.C. recommendations. Thus, if used in conjunction with suitable B & K microphones, the 2120 can be used as either a precision sound level meter or a precision impulse sound level meter. Shock and vibration can also be measured over wide dynamic and frequency ranges when the 2120 is used with one of the B & K vibration transducers. Its peak indication is particularly useful in this instance for the measurement of short duration shocks down to 20 microseconds.

Sound analysis and recording is claimed as another use for the 2120 and this is simplified by the provision of a flexible shaft for connection to a B & K level recorder so that the frequency sweep can be driven continuously. In this manner amplitude/frequency spectrograms can be obtained on preprinted frequency calibrated charts. The 2120 can also be used for distortion measurements down to 0.1% with the filters switched to the bandstop mode.

In spite of its sophistication, the 2120 is claimed to be practically foolproof to operate. The calibration procedure is simple and is aided by the provision of a built-in 1kHz reference oscillator. Once calibrated the instrument is extremely stable over very long periods. Seven calibrated interchangeable meter scales are provided with each instrument and these cover almost every possible application. B & K Laboratories Ltd, Cross Lances Road, Hounslow, Middlesex.

WW 321 for further details.

Precision graticules

Chequers Engraving specializes in the production of high-quality, close-tolerance graticules for oscilloscope and instrument applications. The products feature fine-line techniques (for filled or unfilled use) down to 0.005in line width, and cumulative tolerances over 6in are maintained to within 0.002in. Graticules and separate or associated filters are manufactured from Oroglass, which has been found to be the only proprietary brand capable of sustaining the required high standards of surface finish. Chequers Engraving Ltd, 10 Christina Street, London EC2.

WW 311 for further details

Electrical/electronic aerosols

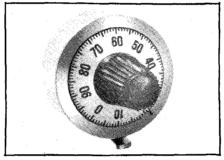
New from PBRA is a range of aerosols with applications in the electrical/electronics industries. The aerosols — part of PBRA's Crown range — include switch and contact cleaners; an anti-static spray; a magnetic tape head cleaner; insulating

varnishes and an inner electric motor cleaner. Also in the range is an aerosol called Electrical 88, which deposits a hydrophobic film — displacing and repelling moisture. The film will not emulsify or harden, will not flake and is non-staining. Effective from -10° C to $+204^{\circ}$ C, Electrical 88 conforms to MIL-C-81309(WP). PBRA Ltd, 33 Holmethorpe Avenue, Redhill, Surrey.

WW 302 for further details

Locking dial

R. C. Knight have introduced a new design of single turn locking dial incorporating the "Digitool" collet mounting system which ensures a positive lock to component shaft without the use of a grub screw. Designated type 25B this dial has an o.d. of 1in and can be used on



several spindle sizes up to a maximum of $\frac{1}{4}$ " dia. Standard colours are silver satin and black and other colours are available on request. Prices range from 90p each and deliveries in the majority of cases can be made from stock. R. C. Knight Ltd, 20 Solent Avenue, Lymington, Hampshire SO4 9SD.

WW 320 for further information.

Audio tape

In response to changing recording techniques EMI Tape Ltd have added a mattbacked tape with a dynamic range of 76.5dB to their range of professional recording tapes. Known as Emitape 817 it can be used on sideless tape hubs or single sided spools. A polyester base with a matt back makes high-speed spooling and uniform wind possible even when using hubs or single-sided spools. Amplitude modulation noise level does not exceed 38.5dB during operation at both 38.1cm/s (15in/s) and 19.05cm/s (7.5in/s) recording speeds.

The tape is available in four standard widths 6.25mm ($\frac{1}{4}$ in), 12.70mm ($\frac{1}{2}$ in), 25.4mm (1in) and 50.8mm (2in), for all mono and stereo recording requirements at 38.1cm/s (15in/s) and 19.05cm/s (7.5in/s). It is supplied in lengths up to 1000m, and has an ultimate tensile strength of 210N/mm² (4.2kg/ $\frac{1}{4}$ in width) with a yield strength (F3) of 110N/mm² (2.3kg/ $\frac{1}{4}$ in width). The nominal overall thickness of the new tape is 56μ m (with a nominal magnetic coating thickness of 18μ m).

At 38.1cm/s (15in/s) the tape's dynamic range is 76.5dB, for full track width, and 72.5dB for a stereo track. Signal-to-print, level at both recording speeds is 56.5. EMI Tape Ltd, Blyth Road, Hayes, Middlesex. WW 312 for further details

Digital electrometer

Keithley Instruments has just introduced a new digital electrometer, the model 616. It features voltage sensitivity from 10V per digit to 200V with an input resistance of greater than $2 \times 10^{14} \Omega$. The instrument can be ranged either automatically, manually, or remotely (with an optionally available digital output/control).

Current sensitivity of the model 616 is $10^{-16}\mathrm{A}$ per digit to $100\mathrm{mA}$. In addition to current and voltage it is also capable of measuring resistance (up to $2\times10^{14}\Omega$), and it will also integrate current (measure charge) from 10^{-15} coulomb per digit to 10^{-5} coulomb. The 6162 outputs are fully isolated and provide complete data, timing outputs, and strobe lines. The open collector b.c.d. outputs are fully compatible with most t.t.l. and d.t.l. circuits presently used on printers and computers. Keithley Instruments Ltd, 1 Bolton Road, Reading, Berks RG2 ONL.

WW 304 for further details

Unwrapping tool

As part of the O.K. Machine and Tool Corporation range of wire wrapping equipment, Vero Electronics are now able to offer deliveries from stock of a speed unwrap tool. The SU2026 unwrapping bit with the G100/R3394 speed unwrap tool has been designed to remove conventional right hand solderless wire wrap connections from 20 to 26 gauge wire without damage to the terminal post. By using the SU2832 bit 28 to 32 gauge wire is removed. The tool is specially useful where a large number of connections require removal in the minimum time. Vero Electronics Ltd, Industrial Estate, Chandler's Ford, Eastleigh, Hampshire.

WW 305 for further details

Trimming capacitors

Jackson Brothers (London) announce a a new range of miniature trimming capacitors, which have a thickness of only 2.8mm. The base of the trimmer is a glass fibre board 10.5 × 8mm. Terminations can be made direct to the board, alternatively pins are supplied so that the user may mount the device on to printed circuit boards, the mounting centres are 5.1mm (0.2m). The nominal capacitance ranges offered are from 2.5 to 10pF, 3.5 to 14pF, 6 to 25pF and 8 to 40pF. Operating temperature is from -20° C to $+70^{\circ}$ C and the devices have a breakdown voltage of 1.5kV at maximum capacitance. Jackson (London) Ltd, Kingsway, Waddon, Croydon, Surrey CR9 4DG.

WW 306 for further details

Seen at the London Components Show

Character generator

Motorola have introduced the MCM6571L, claimed to be one of the most comprehensive read-only memory character generators in a standard range, and using Motorola's n-channel metal gate m.o.s. process to ensure direct c.m.o.s. and t.t.l. compatibility and high operating speed.

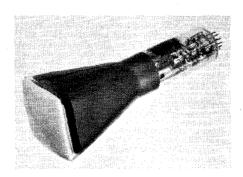
The MCM6571L contains 128 characters, each character being formed by a matrix of seven horizontal and nine vertical dots. Each character is addressed by seven input lines in a modified USASCII code. Once a character has been addressed, the individual rows of seven bits appear on seven output pins under control of a four-bit row select input. Address access and row select times are typically 350ns (500ns max).

In addition to the 8192 bits of memory used to store symbols and characters, the MCM6571L incorporates a 128-bit matrix which controls character position automatically. Certain 'characters normally extend below the base line (j, y, p and q for example) and the 128-bit matrix causes these characters to appear in their correct positions relative to the other characters. Each 7×9 character is in fact stored in one of two positions in a 7×16 matrix (the lower position allowing portions of characters to appear below the base line). The 128-bit matrix automatically informs the addressing circuitry if the character is in the upper or lower position, thereby replacing a substantial amount of external circuitry. Housed in a 24-lead package, the MCM6571L has a typical input capacitance of 4pF and dissipates 600mW. Motorola Semiconductors Ltd, York House, Empire Way, Wembley, Middlesex. WW353 for further details

Cathode ray tubes

New tubes from Brimar include: a 7cm diagonal rectangular instrument tube, the D7-201 (see photograph) which is an improved version of the D7-200. The improvements include modified glassware and higher y-plate sensitivity. The suitability of this tube for low power transistorized equipment has been improved by the introduction of a low wattage heater (6.3V, 0.12A).

The requirement to present high densities of information on large area displays and the need to minimize the overall



length of data terminals have led to the development of the M38-140. This rectangular tube of 38cm diagonal uses 110° deflection and has a neck diameter of 28mm.

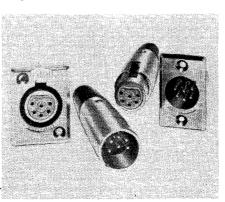
The D10-230 is a 10cm round, single accelerator tube intended for low-cost oscilloscopes and waveform monitoring applications. Features include a flat face, small neck diameter and overall length, 80×64 mm useful screen area and a low deflection voltage requirement. Thorn Radio Valves & Tubes Ltd, Mollison Avenue, Brimsdown, Enfield, Middlesex EN3 7NS.

WW357 D7-201 WW358 M38-140 WW359 D10-230

Audio connectors

F. W. O. Bauch, as sole U.K. distributor, displayed selections from the Switchcraft Q-G range of audio connectors, featuring a streamlined design, rugged construction and simple snap-in positive connection. They are fully compatible with other leading makes, permitting ready changeover and/or extension.

Included in the range is a connector adaptor with a male insert at one end and



a female insert at the other. With 3, 4, or 5 contacts, this connector is particularly designed for use in situations requiring a transformer, attenuator circuit or other electronic circuitry in line with a microphone input.

Also available is the series of cord plugs for direct connection with microphones. The slide switch, in-built into the connector shell, permits finger-tip switching control of microphones and in-line audio and control circuits. The cord plug incorporates a shielding feature which protects the mated shell, thereby making all contacts available for control of audio circuits. F. W. O. Bauch, Ltd, 49 Theobald Street, Boreham Wood, Herts WD6 4RZ.

WW368 for further details

Varicap TV tuner

New from AB Electronic Components is a thick film v.c.d. tuner for television receivers. The tuner is claimed to be the first in Europe using thick film techniques, and possibly the first in the world. It is of small size, $65 \times 34 \times 17$ mm, and offers a British second source supply for Continental products. Complete compatibility with these other products is given by special adaptor printed circuit boards which, if necessary, ensure correct supply voltages and pin connections. AB Electronic Components Ltd, Abercynon, Glamorgan. WW364 for further details

Transmitter transistor

Among new transistors developed by Mullard for use in transmitters is the BLX95. This is an n-p-n, silicon, planar, epitaxial device that can be used under class A, B or C conditions with a supply voltage of 28V. It has a capstan envelope with its leads isolated from the mounting stud. In an un-neutralized, commonemitter, class B circuit, a typical performance for the BLX95 under c.w. input conditions would be:

 $\begin{array}{lll} V_{cc} & -- 28 \mathrm{V} \\ f_T & -- 470 \mathrm{MHz} \\ P_L & -- 40 \mathrm{W} \\ I_C & -- 2.4 \mathrm{A} \\ \mathrm{Efficiency} & -- 60 \% \end{array}$

Mullard Ltd, Mullard House, Torrington Place, London WC1E 7HD.

WW 361 for further details

25MHz dual trace scope.

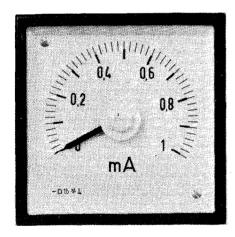
A new addition to the Scopex range of oscilloscopes is a solid-state dual trace version, the 4D-25 at £175 with 3% measuring accuracy on both vertical and horizontal deflection systems. Bandwidth is d.c. to 25MHz, but more significantly, 25MHz measured at full screen can be displayed over the total screen area. Signal delay is operative on both vertical channels and permits the leading edge of

a displayed waveform to be investigated. The timebase has 19 ranges from 200ns/cm to 200ms/cm plus a ×5 magnifier to increase the fastest range to 40ns/cm. Trigger level and polarity are both selected on one control and in the absence of a trigger signal the 4D-25 goes to bright line auto. A tunnel diode stage is used for reliable stable triggering over the full range of sweep speeds. The dimensions of the 4D-25 are 15.3cm (height) × 31.2cm (width) × 43.5cm (depth). Scopex Instruments Ltd, Pixmore Industrial Estate, Pixmore Avenue, Letchworth, Herts SG6 1.II.

WW360 for further details

Long scale meters

The range of meters by British Physical Laboratories included moving-coil panel versions, hermetically sealed meters, long-scale meters, and switchboard meters, also including three new Europa types — the E31G panel meter, available both for panel and sub-panel mounting, with 95° arc movement in a small barrel attractive rectangular case assembly. The E72F (in photo) and E96F square Fulscale meters, with 250° arc movement,

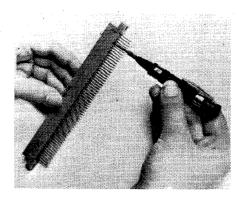


are in standard DIN-size cases. Other new products included non-linear (logarithmic) moving-coil movements, a large dual movement 270° arc switchboard meter being designed for use on the Advanced Passenger Train, Fulscale meters with illumination designed for the boating instrumentation market, and two sizes of flat stackable edgewise meters. British Physical Laboratories, Radlett, Herts WD7 7HJ.

WW367 for further details

Edge connectors

A new edge connector by Carr Fastener is available with up to 90 contacts, single or double sided, on a 0.1in, 2.54mm pitch. The one piece single cantilever contact is a new Cinch design with a narrow face width ensuring low insertion force characteristics over the full printed circuit board thickness range of 0.054in to 0.072in. A firm even pressure is claimed to provide optimum conductivity while



ensuring minimum wear on the foil and a rigid wiring terminal of 0.63mm square section allows for mini-wrap or solder wiring.

A feature offered is the ease with which contacts can be removed and replaced. Removal can be done without disturbing wiring, even when connectors are mounted closely side-by-side.

The polyester moulding is designed to resist all known solvents and has suitable temperature properties to permit solder wiring. Moulded brackets are ultrasonically welded to the body to give the advantages of integral moulded fixings. Working voltage is 600V d.c. or a.c. and current rating is 5A d.c. and a.c. Contact identification and polarization keys are optional. Carr Fastener Company Ltd, Nottingham Road, Stapleford, Nottingham NG9 8AJ.

WW363 for further details

Microwave p-i-n diodes

Intended for 50-ohm broadband stripline circuits, new microwave diodes by AEI Semiconductors, types DC2418A and DC2419A offer a choice of bias polarity and both switch in 50ns. Capable of handling 30 watts c.w., they have an isolation of 25dB at 10GHz and are useful to 20GHz with an insertion loss of 0.5dB. Although having a higher insertion loss, 1dB, and a lower isolation, 20dB, the DC2412A switches in 5ns while handling 12 watts c.w. AEI Semiconductors Ltd, Carholme Road, Lincoln LN11SG. WW 352 for further details

Large-screen oscilloscopes

Lan-Electronics have announced the availability of two new large-screen oscilloscopes, intended for demonstration, education, medical investigation and any other applications requiring large, low-frequency display. The display units themselves are either 20inch or 11inch 110° tubes, with either green long-persistence or white short-persistence traces, and are designed to work in conjunction with local or remote control preamplifiers, which offer four channel operation. Additionally, the display units can be used as self-con-

tained instruments and flexibility is enhanced by the t.t.l.-compatibility of their interfaces. Bandwidth in the y direction is 0-45kHz at -3dB at a sensitivity of 67.5 mV/cm. The x amplifier covers 0-15kHz (-3dB) at 50mV/cm and z modulation is directly coupled to 4.5MHz, requiring 4.5V for full modulation. The Model A control unit, the first of a series, offers a four-channel y preamplifier and triggered timebase providing a maximum sweep speed of 20µs/cm. An unusual feature of the y channel switching circuitry is that all channels may be displayed in turn by the use of a "mechanical trigger" input. Lan-Electronics Ltd, 95 Farnham Road, Slough, SL2 3AF, Bucks.

WW351 for further details

Digital tester

The Wayne Kerr "Swift" tester for digital p.c.bs compares the performance of two boards – one of which is known – and checks for any non-parity of outputs for all combinations of up to 27 inputs. It accepts double or single-sided boards with

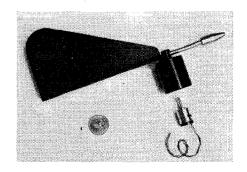


up to 64 ways on the edge connector. Programming is by patch-panel for experimental testing or short production runs. A pre-wired matrix plug is used for longer test runs. Wayne Kerr Co Ltd, Durban Road, Bognor Regis, Sussex PO22 9RL.

WW 356 for further details

Wind vane potentiometer

One of the features displayed by May Precision Components was a wind direction indication system for marine and similar applications. The Deswynn System, which is one of the commercial designs available for yachting and in-shore shipping, is a d.c. system in which a mast-mounted wind vane provides direct drive to a lightweight potentiometer. As



the vane reacts to each change of direction, so the potentiometer reflects this on a cockpit console. The potentiometer MR 101 was designed specifically for this application, and features low-torque, low-weight factor, high linearity and is environmentally sealed. May Precision Components Ltd, Bowlers Croft, Honywood Road, Basildon, Essex.

WW 366 for further details

Counter frequency divider

AMF Venner have produced a 500MHz non-displaying frequency divider for use with their own "77" range of counters and other types capable of measuring frequencies of 5MHz and over. The input to the unit is divided in frequency by factors of 100 and 10, the two outputs being provided simultaneously at two front-panel sockets. Signals of 10mV r.m.s. into 50 ohms are acceptable, and an indication is given when the input is greater than 20mV. The output to the counter is a 500mV p.p. square wave when terminated by a normal counter input circuit. A built-in mains power supply is provided. AMF Venner Ltd, Kingston by-pass, New Malden, Surrey.

WW 355 for further details

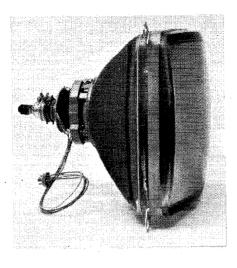
High stability signal generator

A synchronized signal generator, the Marconi TF2006/1, provides locking over the frequency range 10 to 500MHz in 100Hz steps with all possibilities of false locking removed. Seven front panel switches are used to set up the required frequency and a separate two-position switch is set to the "tune" position. The frequency is then adjusted so that two front-panel lamps are both flashing and with the switch then set to the "lock" position the generator is automatically locked to the required frequency with no possibility of ambiguity. With digital synchronizer incorporated, TF2006/1 renders unnecessary the use of an external counter with associated leakage problems at low output levels. Marconi Instruments Ltd, Longacres, St. Albans, Herts AL4

WW 362 for further details

Colour TV tube

An unusual colour television cathode-ray tube shown by Thorn Colour Tubes Ltd has the deflection, convergence and purity components supplied integral with the tube and already set up — the idea being to eliminate the convergence and purity setting-up operations that are normally necessary during receiver manufacture, subsequent tube replacement or receiver servicing. Two other features which distinguish it from the conventional shadow-mask tube are a vertically striped



three-colour phosphor screen and corresponding slotted shadow-mask, and a horizontal in-line electron gun. Known as the Precision In-line Colour TV Tube assembly, the device appears to be similar to an RCA tube described in our October 1972 issue (p.466). It has a 51cm (diagonal) screen, a 90° deflection angle and a 29mm neck diameter, the overall length of the tube being 417mm. Designated A51-160X, the colour tube will be available in the Autumn of 1974. Thorn Colour Tubes Ltd, Mollison Avenue, Brimsdown, Enfield, Middx.

WW 365 for further details

Digital voltmeter

A new digital voltmeter, the Advance DVM44, is a small multi-range integrating d.c. digital instrument incorporating a 3½-digit display with automatic polarity indication. Analogue to digital conversion is achieved using a dual-slope integration system. Five ranges extend from 0 (to 199.9mV, 1.999V, 19.99V, 199.9V) to 1000V, and power input requirements are 100-125V or 200-250V at 45-65Hz. Advance Electronics Ltd, Raynham Road, Bishop's Stortford, Herts.

WW 350 for further details

I.C. logic checker

A new i.c. logic checker by Jermyn Manufacturing gives immediate display of the logic state of each pin in a d.t.l. or t.t.l. dual-in-line integrated circuit. It replaces laborious alternative methods of testing, for example with a single probe and cathode ray tube. The checker $-2\frac{1}{2}$ inches long and weighing just $1\frac{1}{2}$ oz – is clipped over a 5V i.c. (up to 16 pins) and monitors and displays the logic state of each terminal on an array of 16 light emitting diodes. A lamp on means logic state 1 or open circuit on that pin, a lamp off means logic state 0 (or ground), and unused terminals and the 5V supply pins also show logic state 1. The checker takes its power - only 200mA with all lamps on from the two pins on the i.c. carrying the 5V supply, and gives an identical reading whichever way round it is clipped. It is suitable for most t.t.l. and d.t.l. devices and is supplied with 24 masks to help check the most common logic configuration. An i.c. insertion tool is included in the kit. Jermyn Manufacturing, Vestry Estate, Sevenoaks, Kent.

WW 354 for further details

Precision resistors in TO-5 cans

Originally developed for military use, Alma type SN precision wirewound resistors will be in production later this year. The resistors feature bobbinless construction, all-welded terminations, are housed in TO-5 cans and meet nearly all military specifications. Values, in the range ten ohms to two megohms, can be set to 0.01% with a long-term stability of 50 parts per million (0.005%) per year, reducing to 20 parts per million after the first year. Rated for watt dissipation at 70°C, temperature coefficient is nominally 10 parts per million, with 2 and 5 parts to special order. Alma Components Ltd, Park Road, Diss, Norfolk IP22 3AY.

WW 372 for further details

Yagi array for 1500MHz

Thought to be the cheapest 1500MHz aerial available for point-to-point links, the J Beam 7127 is unusual in adopting a Yagi construction. Conventional horn or dish aerials normally require mounting on towers because of wind loading, but using a Yagi array means the aerial can be pole mounted (this model having a wind loading of about 40 lb). Over the band 1450 to 1535MHz (used in the main service industries for point-to-point communication) the aerial, using the J Beam slot-fed radiator with 18 parasitic elements, shows a gain of 16dB with less then 0.5dB variation and with a v.s.w.r. of between 1.2 and 1.3. Front-to-back ratio is 26dB and half-power angle is 26° in the E-plane. It measures five feet in length and the 5-in dia. glass fibre "radome" gives protection against icing. Cost is around £100. J Beam Engineering Ltd, Rothersthorpe Crescent, Northampton.

WW 369 for further details

Accurate sine/cosine potentiometers

Using a conductive plastics material, Penny & Giles achieve a very accurate sine/cosine law in their latest potentiometers. Designed primarily for providing range circles in radar displays, they conform to a sinusoidal law to within 0.1% pk-pk. They are available in six values from 1 to $10 \text{k}\Omega$ at $\pm 10\%$ tolerance. Other conformity values of 0.5 and 1% are also available. Penny & Giles Conductive Plastics Ltd, Blackwood, Mon.

WW 370 for further details

Save millimetres AND money with this new **Brimar Tube**

If you're planning an oscilloscope - or any instrument for waveform monitoring - Brimar has good news for you. The Brimar D10-230 mono-accelerator tube is a real winner where low cost and small size are important considerations.

Short in length, small around the neck, it can save you a whole lot of space . . . give you far greater freedom in your designing. The flat 4" diameter face offers minimum total scans of 80mm and 64mm, in X and Y directions respectively. And its voltage requirements are within the capability of normal, low-cost, transistorised circuits.

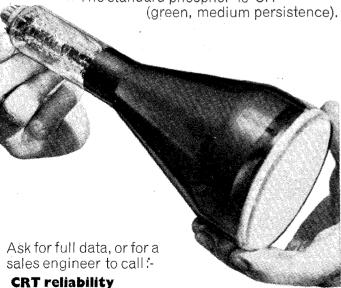
This tube with its Standard B14G Colour tube base also allows economies to be made in Socket costs and, being round, requires no twist coil.

- ★ Maximum overall length: 260mm
- ★ Maximum neck diameter: 38mm
- ★ First anode voltage: Va1, 1500V
- ★ Second anode voltage for focus: Va2, 230 to 380V
- ★ Grid voltage for spot cut-off (approx.):

Vg −30 to −65V

★ X plate deflection coefficient: Dx, 21 to 26V/cm ★ Y plate deflection coefficient: Dy, 13 to 16V/cm

★ The standard phosphor is GH





Thorn Radio Valves and

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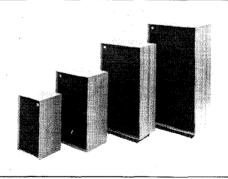












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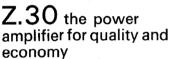
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Sinclair Project 60

Now-the Z.50 Mk.2

with built-in automatic transient overload protection

When originally introduced, the Sinclair Z.50 proved how it was possible to design and produce a popularly priced modular power amplifier having characteristics to challenge the world's costliest amplifiers. Many thousands of Z.50's are now giving excellent service day in, day out. But we have also learned that constructors do not always use their Z.50's ideally. That is why we have introduced modifications whereby risk of damage through mis-use is greatly reduced and performance further enhanced. The Z.50 Mk.2 has improved thermal stability, more accurately regulated D.C. limiting to ensure more symetrical output voltage swing and clipping and still less distortion at lower power. Z.50 Mk.2 is compatible with all other Project 60 modules, and may be incorporated to advantage in existing systems. Eleven silicon epitaxial planar transistors are now used, two more than in the original Z.50; circuitry has been re-designed, making this versatile high performance amplifier better than ever.





with free manual

The Z.30 provides excellent facilities for the constructor requiring a high fidelity audio system of less power than that available from Z.50's. Using a power supply of 35 volts, Z.30 will deliver 15 watts RMS into 8 ohms, or 20 watts RMS into 3 ohms using 30 volts. Total harmonic distortion is a fantastically low 0.02% at 15 watts into 8 ohms with signal to noise ratio better than 70 dB unweighted. Input sensitivity 250mV into 100K ohms. SIze 80 x 57 x 13 mm ($3\frac{1}{6}$ x $2\frac{1}{4}$ x $\frac{1}{2}$) Z.30, Z.50 and Z.50 MK.2 modules are compatible and interchangeable.

Guarantee

If, within 3 months of purchasing any product direct from Sinclair Radionics Ltd., you are dissatisfied with it, your money will be refunded at once. Many Sinclair appointed Stockists also offer this same guarantee in co-operation with Sinclair Radionics Ltd.

Sinclair Radionics Ltd.

Each Project 60 module is tested before leaving our factory and is guaranteed to work perfectly. Should any defect arise in normal use, we will service it at once and without any charge to you, if it is returned within two years from the date of purchase. Outside this period of guarantee a small charge (typically £1.00) will be made. No charge is made for postage by surface mail. Air Mail is charged at cost.



Brilliant new technical specifications

Input impedance 100 K Ω Input (for 30w into 8Ω) 400mV Signal to noise ratio, referred to full o/p at 30v HT 80dB or better Distortion 0.02% up to 20W at 8Ω . See curve

Distortion 0.02% up to 20W at 8 Ω . See curve **Frequency response** 10Hz to more than 200 KHz \pm 1dB

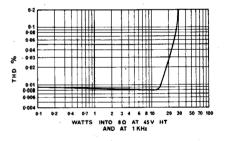
Max. supply voltage 45v (4 Ω to 8 Ω speakers) (50v 15 Ω speakers only)

Min. supply voltage 9vLoad impedance — minimum: 4Ω at 45v HT Load impedance — maximum: safe on open

ircuit



with free manual £5.48



Typical Project 60 applications

| System | The Units to use | together with | Units cost | |
|---|---|--|------------|--|
| Simple battery record player | Z.30 | Crystal P.U., 12V battery volume control, etc. | £4.48 | |
| Mains powered record player | Z.30, PZ.5 | Crystal or ceramic P.U. volume control, etc. | £9.45 | |
| 12W. RMS continuous sine wave stereo amp. for average needs | 2 x Z.30s, Stereo 60; PZ.5 | Crystal, ceramic or mag. P.U., F.M. Tuner, etc. | £23.90 | |
| 25W. RMS continuous sine wave stereo amp. using low efficiency (high performance) speakers | 2 x Z.30s, Stereo 60; PZ.6 | High quality ceramic or magnetic P.U., F.M. Tuner, Tape Deck, etc. | £26.90 | |
| 80W. (3 ohms) RMS continuous sine wave de luxe stereo amplifier. (60W. RMS into 8 ohms) | 2 x Z.50s, Stereo 60; PZ.8, mains transformer | As above | £34.88 | |
| Indoor P.A. | Z.50, PZ.8, mains transformer | Mic., guitar, speakers, etc., controls | £19.43 | |

the world's most advanced high fidelity modules

Stereo 60 Pre-amp/control unit



Designed specifically for use on Project 60 systems, the Stereo 60 is equally suitable for use with any high quality power amplifier. Since silicon epitaxial planar transistors are used throughout, a really high signal-to-noise ratio and excellent tracking between channels is achieved. Input selection is by means of press buttons, with accurate equalisation on all input channels. The Stereo 60 is particularly easy to mount.

SPECIFICATIONS—Input sensitivities: Radio – up to 3mV. Mag. p.u. 3mV: correct to R.I.A.A. curve ±1dB:20 to 25,000 Hz. Ceramic p.u. – up to 3mV: Aux – up to 3mV. Output: 250mV. Signal to noise ratio: better than 70dB. Channel matching: within 1dB. Tone controls: TREBLE+12 to —12dB at 10KHz: BASS +12 to —12dB at 100Hz. Front panel: brushed aluminium with black knobs and controls. Size: 66 x 40 x 207mm.

Built, tested and guaranteed.

£9.98

Project 60 Stereo F.M. Tuner



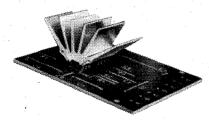
The phase lock loop principle was used for receiving signals from space craft because of its vastly improved signal to noise ratio. Now, Sinclair have applied the principle to an F.M. tuner with fantastically good results. Other advanced features include varicap diode tuning, printed circuit coils, an I.C. in the specially designed stero decoder and switchable squelch circuit for silent tuning between stations. In terms of a high fidelity this tuner has a lower level of distortion than any other tuner we know. Stereo broadcasts are received automatically, a panel indicator lighting up as the stereo signal is tuned in. This tuner can also be used to advantage with most other high fidelity systems.

SPECIFICATIONS—Number of transistors: 16 plus 20 in I.C. Tuning range: 87.5 to 108MHz. Sensitivity: $7\mu V$ for lock-in over full deviation. Squelch level: Typically $20\mu V$. Signal to noise ratio: >65dB. Audio frequency response: 10Hz - 15KHz (± 1 dB). Total harmonic distortion: 0.15% for 30% modulation. Stereo decoder operating level: $2\mu V$. Cross talk: 40dB. Output voltage: 2×150 mV R.M.S. maximum Operating voltage: 25-30VDC. Indicators: Stereo on; tuning. Size: $93\times 40\times 207$ mm.

Built and tested. Post free.

£25

Super IC.12 Integrated circuit high fidelity amplifier



Having introduced Integrated Circuits to hi-ficonstructors with the IC.10, the first time an IC had ever been made available for such purposes, we have followed it with an even more efficient version, the Super IC.12, a most exciting advance over our original unit. This needs very few external resistors and capacitors to make an astonishingly good high fidelity amplifier for use with pick-up, F.M. radio or small P.A. set up, etc. The free 40 page manual supplied, details many other applications which this remarkable IC. make possible. It is the equivalent of a 22 transcript of the supplications of the supplicat

sistor circuit contained within a 16 lead DIL package, and the finned heat sink is sufficient for all requirements. The Super IC.12 is compatible with Project 60 modules which would be used with the Z.50 and Z.30 amplifiers. Complete with free manual and printed circuit board.

SPECIFICATIONS

Output power: 6 watts RMS continuous (12 watts peak). 6–8Ω. Frequency Response: 5 Hz to 100KHz±1dB. Total Harmonic Distortion: Less than 1%. (Typical 0-1%) at all output powers and frequencies in the audio band (28V). Load Impedance: 3 to 15 ohms. Input Impedance: 250 Kohms nominal. Power Gain: 90dB (1,000,000,000 times) after feedback. Supply Voltage: 6 to 28V. Quiescent current: 8mA at 28V. Size: 22×45×28mm including pins and heat sink.

Manual available separately 15p post free,

With FREE printed circuit board and 40 page manual.

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Power Supply Units The new PZ.8 Mk.3

The most reliable power supply unit ever made available to constructors. Brilliant circuitry makes failure from over load and even direct shorting of the output impossible. This is due to an ingenious re-entrant current limiting principle which, as far as we know has never before been available in any comparable unit outside the most expensive laboratory equipment. Ripple and residual noise have been reduced to the point of almost total elimination. This is, of course, the perfect unit for Project 60 assemblies, particularly where the new Z.50 MK.2 amplifiers are used. Nominal working voltage – 45.

PZ.8 Mk.3—£7.98

(Mains transformer, if required) £5.98

PZ.5 30v. unstabilised

(not suitable for Project 60 tuner) £4.98

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(not suitable for IC. 12) £7.98

Project 605

the easy way to buy and build Project 60 without soldering



Project 605 in one pack contains: one PZ.5, two Z.30's, one Stereo 60 and one Masterlink, which has input sockets and output components grouped on a single module and all necessary leads cut to length and fitted with clips to plug straight on to the modules thus eliminating all soldering.

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| AC188 0-24 | BC107 0:10 BC108 0:10 | BC209 0·13 BC212L 0·12 | BF127 0.55 BF152 0.61 | BSY29 0.17 BSY38 0.20 | OC202 0.31 OC203 0.28 | | 0·16 2N3010 0·77 0·33 2N3011 0·16 | 2N3906 0·30 2N4058 0·13 | 40362 0.50 |
| AC188K 0.22 ACY17 0.28 | BC109 0·11 | BC213L 0·12 | BF153 0.50 | BSY39 0.20 | OC204 0.28 | 2N929 (| 0.23 | 2114000 0110 | 1 |
| ACY18 0.22 | BC113 0·11 | BC214L 0·16 | BF154 0.50 BF155 0.77 | BSY40 0.31 BSY41 0.31 | OC205 0·39 | |)·23)·22 | DIODES AND RECTIF | TIERS |
| ACY19 0 22 ACY20 0 22 | BC114 0·17 BC115 0·17 | BC225 0.28 BC226 0.39 | BF156 0 53 | BSY95 0:14 | OC309 0.44 OCP71 0.48 | | 0.24 AA119 0.08 | BY133 0.23 | OA10 0-39 |
| ACY21 0 22 | BC116 0.17 | BCY30 0.27 | BF157 0.61 | BSY95A 0.14 | ORP12 0.48 | | D-16 AA120 O-09 D-16 AA129 O-09 | | OA47 0.08 OA70 0.08 |
| ACY22 0 18 ACY27 0 20 | BC117 0·17 BC118 0·11 | BCY30 0.27 BCY31 0.29 | BF158 0.61 BF159 0.66 | Bul05 2-20 ClllE 0-55 | ORP60 0.44 ORP61 0.44 | | 0·19 AAY30 0·10 | 0.46 | OA70 0.08 OA79 0.08 |
| ACY28 0 21 | BC119 0.33 | BCY32 0.33 | BF160 0.44 | C400 0.33 | P346A 0.22 | . 2N1305 (| 0·19 AAZ13 0·13 | | OA81 0.08 |
| ACY29 0 39 ACY30 0 31 | BC120 0-88 BC125 0-13 | BCY33 0.24 BCY34 0.28 | BF162 0.44 BF163 0.44 | C407 0.28 C424 0.22 | | | D-23 BA100 0-13 D-23 BA116 0-23 | | OA85 0·10 OA90 0·07 |
| ACY31 031 | BC126 0.20 | BCY70 0.16 | BF164 0 44 | C425 0·55 | ST140 0.14 | 2N1308 (| D-26 BA126 0-24 | BYZ13 0.28 | OA91 0-07 |
| ACY34 0 23 ACY35 0 23 | BC132 0·13 BC134 0·20 | BCY71 0.20 BCY72 0.16 | BF165 0.44 . BF167 0.24 | C426 0:39 C428 0:22 | ST141 0.19 TIS43 0.88 | | 0·26 BA148 0·16 0·22 BA154 0·18 | BYZ16 0.44 BYZ17 0.39 | OA95 0.08 OA200 0.07 |
| ACY36 031 | BC135 0.13 | BCZ10 0.22 | BF173 0.24 | C441 0.33 | UT46 0.30 | 2N1711 (| D-22 BA155 0-16 | BYZ18 0.39 | OA202 0.08 |
| ACY40 0 19 | BC136 0·17 | BCZ11 0.28 | BF176 0.34 BF177 0.39 | C442 0.33 | | | 0.35 BA156 0.18 0.50 BY100 0.17 | BYZ19 0.31 CG62 | SD10 0.06 SD19 0.06 |
| ACY41 0 20 ACY44 0 39 | BC137 0-17 BC139 0-44 | BCX12 0.28 BD121 0.66 | BF178 0.33 | C444 0.39 C450 0.24 | 2G302 0·21 2G303 0·21 | | 0·41 BY101 0·1 3 | (OA91 Eq.) | 1N34 0.8 |
| AD130 0 42 | BC140 0·33 | BD123 0.72 | BF179 0·33 | MAT00 10.21 | . 2G304 0·27 | 2N2147 (| 0.79 BY105 0.19 | 0008 | 1N34A 0.08 |
| AD140 0 53 AD142 0 53 | BC141 0.33 BC142 0.33 | BD124 0.66 BD131 0.55 | BF180 0.33 BF181 0.33 | MAT101 0.22 MAT120 0.21 | 2G306 0·44 2G308 0·39 | | 0·63 BY114 0·16 0·66 BY126 0·16 | (OA70-0A79 | 1N914 0.07 1N916 0.07 |
| AD143 0 43 | BC143 0.33 | BD132 0.66 | BF182 0.44 | MAT121 0.20 | 2G309 0·39 | 2N2192 (| 0.39 BY127 0.17 | Eq.) 0.07 | 1N414B 0.07 |
| AD149 0 55 AD161 0 37 | BC145 0.50 BC147 0.11 | BD133 0.72 BD135 0.44 | BF183 0.44 BF184 0.28 | MPF102 0.46 MPF104 0.41 | | | 0·39 BY128 0·17 0·39 BY130 0·18 | | 18021 0·11 18951 0·07 |
| 122101 001 | | | | | | | | | |
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Q26 | 8 OA95 Germanium diodes sub-min. | 1.69 | 0.55 |
Q27 | 2 10A PIV Silicon rectifiers | 18425R | 0.55 |
Q28 | 2 Silicon power rectifiers | 18425R | 0.55 |
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1 × 2N698 | 0.56 | 0.56 |
Q31 | 6 Silicon switch transistors 2N708 NPN | 0.55 |
Q32 | 3 PNF Silicon transistors 2 × 2N113 | 0.55 |
Q33 | 8 Silicon Switch transistors 2N708 NPN | 0.55 |
Q33 | 3 Silicon NPN transistors 2N1711 | 0.55 |
Q34 | 7 Silicon NPN transistors 2N2369, 500MHz |
Q35 | 3 Silicon PNP TO-5, 2 × 2N2904 | 1 × 0.55 |
Q36 | 7 N3646 TO-18 plastic 300 MHz NPN | 0.55 |
Q37 | 3 N3063 NPN Silicon transistors | 0.55 |
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Q38 | 7 NPN transistors 4 × 2N3703, 3 × 2N3702 | 0.55 |
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UTC36 = 12 × 7450
UTS61 = 12 × 7450
UTS61 = 12 × 7450
UTS61 = 12 × 7450
UTS62 = 12 × 7450
UTC60 = 12 × 7460
UTC70 = 8 × 7470
UTC70 = 8 × 7470
UTC72 = 8 × 7470
UTC75 = 8 × 7470
UTC75 = 8 × 7470
UTC75 = 8 × 7470
UTC76 = 8 × 7480
UTC76 = 8 × 7480

All indicators

0.9 + Decimal

point. All side

viewing. Full
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types available on request.

Price

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UIC91 = 5 × 7480
UIC91 = 6 × 7490
UIC91 = 6 × 7491
UIC93 = 5 × 7492
UIC93 = 5 × 7492
UIC93 = 5 × 7494
UIC95 = 5 × 7494
UIC96 = 5 × 7496
UIC102 = 5 × 7496
UIC102 = 5 × 74100
UIC121 = 5 × 74110
UIC151 = 5 × 74110
UIC151 = 5 × 74110
UIC151 = 5 × 74110

UICXI=25 Assorted 74's 1.55

TYPE PA100

Built to a specification and NOT a price, and yet still the greatest value on the market, the PA100 stereo pre-amplifier has been conceived from the latest circuit techniques. Designed for use with the AL50 power amplifier system, this quality made unit incorporates no less than eight silicon planar transistors, two of these are specially selected low noise NPN devices for use in the input stages. Three switched stereo inputs, and rumble and scratch filters are features of the PA100, which also has a STEREO/MONO switch, volume, balance and continuously variable bass and treble controls.



SPECIFICATION:
Frequency response
Harmonic distortion
Inputs: 1. Tape head
Inputs: 1. Tape head
2. Radio, Tuner
3. Magnetic P.U.
All Input voltages are for an output of 250mV.
Tape and P.U. inputs equalised to RIAA curve
within ± 1.lB from 20Hz to 20kHz.

Bass control Treble control Filters: Rumble (high pass) Scratch (low pass) Signal/noise ratio Input overload Supply Dimensions

±15dB at 20Hz 8kHz better than +65dB +26dB

+26dB +35 volts at 20mA 292 × 82 × 35 mm

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only £13.15



The STEREO 20

The 'Stereo 20' amplier is mounted, ready wired and tested on a one-piece chassis measuring 20 cm × 14 cm × 5.5 cm This compact unit comes compile with an information of the control, balance, hase and trible controls. Attractively printed front panel and matching control knobs. The 'Stereo 20' has interfering with the mechanism or, alternatively, into a september of the control knobs. The 'Stereo 20' has control knobs. The 'Stereo 20' has control knobs. The 'Stereo 20' has control knobs of the control knobs

arate cabinet.

Output power 20w peak
Freq. res. 28Hz-28kHz
Harmonic distortion.

typically 0.25% at 1 watt
Treble con. ±14dB at 14kHz

£13.47 free p. & p.

Pak No. Contents UIC00=12±7400 UIC01=12±7401 UIC01=2±7401 UIC02=12±7401 UIC02=12±7403 UIC04=12±7403 UIC04=12±7403 UIC04=12±7405 UIC06=8±7407 UIC01=12±7405 UIC06=8±7407 UIC10=12±7410 UIC13=8±7413 UIC30=12±7420 UIC30=12±7420 UIC30=12±7420 UIC30=12±7420 UIC30=12±7420 UIC30=12±7420 UIC30=12±7420 UIC30=12±7430 UIC Packs cannot be split, but 25 assorted pieces (our mix) is available as PAK UIC X1. NEW COMPONENT PAK RARGAINS

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| C12 | 40 | Paper Condensors professed 4 | 0.55 |
| C13 | 20 | | 0.55 |
| C14 | 1 | | |
| C15 | 4 | | 0.5 |
| C16 | _ | | 0.55 |
| C17 | 20 | | 0.58 |
| C18 | 10 | | 0.55 |
| | 4 | | 0.55 |
| C19 | 3 | | 0.58 |
| C20 | 4 | Sheets Copper Laminate approx. $10'' \times 7''$ | 0.58 |
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| ı | | | UL84 | 0.35 | 3A4 | 0.40 | 6AK5 | 0.35 | 6BJ6 | 0.45 | 6J6 | |
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| ŀ | UABC80 | 0.30 | UY41 | 0.45 | | | 6AL5 | 0·18 | 6BQ7A | 0.43 | 6J7G | 0.30 |
| í | UAF42 | 0.50 | | | 3Q4 | 0.55 | 6AL5W | 0.35 | 6BR7 | 0.85 | 6J7M | 0.35 |
| l | | | UY85 | 0.35 | 384 | 0.33 | 6AM6 | 0.35 | 6BW6 | 0.80 | 6K6GT | 0.55 |
| ł | UBC41 | 0.48 | VR105/30 | 0.35 | 3V4 | 0.48 | 6AN8 | 0.60 | 9BW7 | 0·80 | 6K7 | 0.40 |
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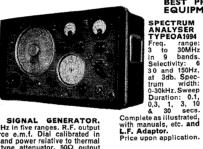
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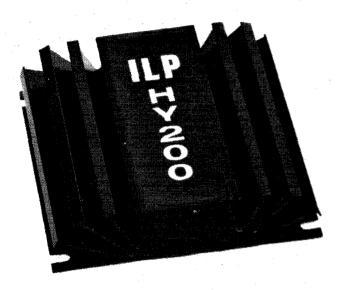
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| 1-10p 1-05p 1-47p 1-37p 44p 41p 1-54p 1-43p | 33μF | 22μг 6½p 47μF 6½p 100μF 8p 150μF 8p 220μF 10p 470μF 13p | 63 VOLT µF 6½p 2-2µF 6½p |
| | IO VOLT | 680μF 20p 1000μF 22p | 1.4·7μF 6 ∔p |
| | 22μF 6½p 47μF 6½p 100μF 6½p 220μF 8p 330μF 10p 470μF 10p | 2200μF 39p 5000μF 68p | 6.8µF 6½p 10µF 6½p 22µF 6½p 68µF 10p 100µF 11p 150µF 13p 220µF 13p |
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Range 10 ohms to 47 meg.
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Range 10— meg ohms

BARGAIN **PACKS**

| Unmarked | |
|-----------------|--|
| Packs | |
| Pack of 25 | |
| | |
| IN <u>4</u> 148 | |
| 55p | |

| Pack o | f IO |
|----------|------|
| | |
| BC108 | 65p |
| BC107 | 65p |
| (Plastic | |

Pack of 10 Plastic BC109 **65p**

| Pack of | f 10 |
|---------|--------|
| BC169 | 65p |
| (unmarl | ced) . |
| hut te | sted |

| 2N2646 |
|------------|
| |
| (unmarked) |
| |
| . 33p each |

| Pack of I | 0 |
|-----------|-----|
| 2N2926G | 65p |
| unbranded | but |
| tested | |

| Unmarke fully te: 2N30 | sted |
|------------------------------|------|
| 1-9 | 33p |
| 10 plus | 27p |

FULLY MARKED TYPES AD161, AD162 M/P

| I-9 | 65p |
|---------|--------|
| IO plus | 60p |
| BC107 | -BC108 |

| BC107-BC BC109 | 108 |
|-------------------|-----|
| 1-9 | 10p |
| 10-99 | 9p |
| 100 plus | 8p |

| ВС | 182L: 212-4 | 3-4- |
|-------------|----------------|------------|
| 1-9 10 p | | llp 10p |

| ACI27 or | ACI28 |
|----------|-------|
| 1-9 | I3p |
| 10 plus | 12p |
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250v.0250v.050mA., 6.3v. 4a., 0.5-6.3v. 3a.
250v.0250v.00mA., 6.3v. 4a., 0.5-6.3v. 3a.
250v.0250v.100mA., 6.3v. 4a., 0.5-6.3v. 3a.
250v.0250v.150mA., 6.3v. 4a., 0.5-6.3v. 3a.
250v.0250v.250mA., 6.3v. 4a., 0.5-6.3v. 3a.
250v.0250v.250mA., 6.3v. 4a., 6.5v. 3a.
250v.0250v.250mA., 6.3v. 4a., 6.3v. 4a.
250v.0250v.250mA., 6.3v. 4a., 6.3v. 4a.
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A full range unit to provide excellent
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| 282812 RCA | 8 TRACK | 19467 EPS | YLON | | .8 TRACK |
| 303489 RCA | 7 TRACK | 76815/948 | BURR | OUGHES | 9 TRACK |
| 73927 RCA | 7 TRACK | 652623DR | 12/10 | 01 | 12 TRACK |
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| 2N457 75 p | AF116 | 25p | 2N3 | 56/OC13 | 9 25 p |
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| BTX-82-300R 300v. | control from | n min. to ma rol fused for (| x. pov | ver using | Triac |
| 26a£2.00 | Phase Cont | rol fused for (| D/L pro | tection Di | men. |
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| McMurdo Red Range, Plu | g RP24 | | | | 56 |
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Airmax Type M1/Y3954 (3 blades) Cast Aluminium alloy impeller & casing (corres-ponds to current type 3965 7½") 230v. 1ph 50c 2900rpm Class "A" insulation

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| Code | Power | Tolerance | Range | Values available | 1 to 9 | 10 to 99 e note belo | 100 up |
|------|-------|-----------------------|---------------------------|---------------------|--------|-------------------------|-----------|
| C | 1/20W | 5% | 82Ω-220ΚΩ | E12 | 9``` | R | 7.5 |
| č | 1/8W | 5% | 4.7Ω-470ΚΩ | E24 | ĭ | ŏ·9 | 0.75 nett |
| č | 1/4W | 5% | $4.7\Omega - 10M\Omega$ | E12 | 1 | 0.9 | 0.75 nett |
| Č | 1/2W | 5% | 4.7Ω -10M Ω | E24 | 1.2 | 1 | 0.9 nett |
| c | 1W | 5% | 4.7Ω -10M Ω | E12 | 2.5 | 2 | 1.6 nett |
| MO | 1/2W | 2% | 10Ω-1ΜΩ | E24 | 4 | 3 | 2 nett |
| ww | 1W | $10\% \pm 1/20\Omega$ | $0.22\Omega - 3.9\Omega$ | E12 | 7 | 7 | 6 |
| ww | 3W | 5% | 1Ω – $10K\Omega$ | E12 | 7 | 7 | 6 |
| ww | 7W | 5% | 1Ω-10ΚΩ | E12 | 9 | 9 | 8 |

VW
 Codes: C = carbon film, high stability, low noise.
 MO = metal oxide, Electrosii TR5, ultra low noise.
 WW = wire wound, Plessey.
 Values:

aues: 12 denotes series: 10, 12, 15, 18, 22, 27, 33, 39, 47, 56, 68, 82 and their decades. 24 denotes series, as E12 plus 11, 13, 16, 20, 24, 30, 36, 1, 51, 62, 75, 91 and their decades.

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2N3055 npn silicon power AC153K pnp germanium low power AC176K npn germanium low power AD161 npn germanium medium power 32p 32p 32p 42p AD161 npn germanium medium power AD162 ppn germanium medium power AF139 ppn germanium UHF BC107—13p; BC108—12p; BC109—13p BC167—11p; BC168—10p; BC169—11p BC177—21p; BC178—1p; BC179—22p BC257—12p; BC258—11p; BC259—13p Standard groupings available. 40p 33p npn

DIODES
OA90, OA91, OA95 each 6p
OA200—9p; OA202—10p Other semi-conductors ACI28—21p AFII7—32p BFY51—19p

> 7403 7404 7405

TTL ICs

Nett Price

20p 20p 20p 25p 25p 25p 25p 20p 20p 20p 20p £1.00 £1.36

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Very many other types listed, described and illustrated in catalogue.

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| 2 way loudspeaker 3 way audio 5 way audio 180° 5 way audio 240° 6 way audio | Socket Socket Socket Socket | 10p 12p 12p 13p | Plug Plug Plug Plug Plug | 12p 12p 15p 15p 15p |
|---|--------------------------------------|--------------------------|--------------------------------------|---------------------------------|
| Lockable types, ph | | | | .sp |

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Prices are in pence each for quantities of the same ohmic value and power rating. NOT mixed values. (Ignore frac-tions of one penny on total value of resistor order.)

clip-on 7
HEATSINK Type 6WI Extruded aluminium 1° C/W, undrilled 60p



TOGGLE SWITCHES

1011C SPST toggle 19p; 409 DPDT toggle 28p. (These are chrome plated, 2.5A rating). 7201 Sub-miniature DPDT 250V a.c./2A 48p



POTENTIOMETER carbon type long spindles. Double wipers for low noise. SINGLE GANG P20 linear 100Ω to $2\cdot2M\Omega$, 12p, JP20 Log, $4\cdot7K\Omega$, to $2\cdot2M\Omega$

1P20 Log, **7/8.2, to 2-21 is 2p. 21p. DUAL GANG linear 4-7KΩ to 2-2MΩ, 42p; Dual gang log, 4-7KΩ to 2-2MΩ, 42p; Log/antilog, 10K, 22K, 47K, 1MΩ only 42p; Dual antilog, 10K only, 42p. Any type with 2A D.P. mains witch 12 navtra.

type with 2A D.P. mains switch, 12p extra. Only decades of 10, 22 & 47 available in ranges quoted. DUAL CONCENTRIC DP20 in any combination of P20 values, 60p; with switch, 72p.

switch, 72p. SLIDER POTS. In values from $4K7\Omega$ to $1M\Omega$, linear or \log , 26p. Escutcheon, white, grey, black, 10p. Knobs, flat, grip type, in 7 colours, 5p each. SKELETON PRE-SETS. Small high quality, type PR linear only: 100Ω , 220Ω , 470Ω , 1K, 2K2, 4K7, 10K, 22K, 47K, 100K, 470K, 1M, 2M2, 5M, $10M\Omega$. Vertical or horizontal mounting, 5p each.

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Nuts 2BA—41p; 4BA—28p; 6BA—26p.
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Screws roundheaded, cheese headed or countersunk.
Other sizes available. Also tags, washers, spacers, etc.

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| ELECIN | Pr | ices in p | ennies | | | | | |
|--------|------|-----------|--------|-----|-----|-----|-----|------|
| μF | 3V | 6.3V | 10V | 167 | 25V | 40V | 63V | 1007 |
| 0.47 | | | | | | | 10 | 7 |
| 1.0 | | | | | | 10 | | 7 |
| 2.2 | | | | | 10 | | 7 | 8 |
| 4.7 | | | | 10 | | 7 | 8 | . 7 |
| 10 | | | | | 7 | 8 | 7 | 8 |
| 22 | | | 7 | | 8 | 7 | 7 | 9 |
| 47 | 7 | | 8 | 7 | 7 | 7 | 9 | 12 |
| 100 | 8 | 7 | 7 | 7 | 7 | 9 | П | 19 |
| 220 | | 8 | 8 | 8 | 9 | 10 | 17 | 27 |
| 470 | 8 | 9 | 9 | 10 | 12 | 17 | 24 | 43 |
| 1000 | 10 | 12 | 12 | 17 | 20 | 24 | 40 | |
| 2200 | . 14 | 17 | 22 | 25 | 36 | 40 | | |
| 4700 | 25 | 28 | 37 | 41 | 54 | | | |
| 10.000 | | 43 | | | | | | |

Smallest size $3.7mm \times 12mm$. Largest size $25.5mm \times 41mm$. Full ranges of many other types of capacitors stocked.

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Radiospares Miniature Maka-

National Plants Plants Plants Switch (in assembly kit form). Shaft 48p. Wafers, MBB—2P5W, IP IIW; BBMIPI2W, 2P6W, 3P4W, 4P3W, 6P2W, each 32p.

Wayechange switches IPI2W, 2P6W, 3P4W, 4P3W, each 24p.





K.30/3 Solid aluminium





38p 20mm. F.17 engraved. 4 in pack
33mm.—two 40p 40
Very many other types in stock—see



Skirt dia



Minitron DIGITAL INDICATOR

TYPE 3015F Seven segment indicator compatible with standard logic modules and power supplies. Figs, 0 9 from well illuminated filament segments to give character of 9mm height plus decimal point. Power requirement 8mA from 5V D.C. per segment. A limited number of alphabetical symbols also available. In 16 lead DIL case ... nett 22-00

Suitable BCD decoder driver type
FLL121T nett
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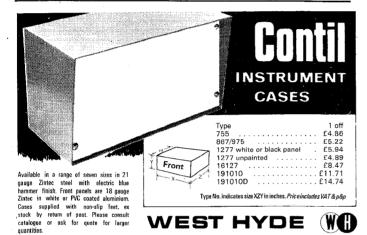
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| 5 Z 3 | 0.45 | 6DT6 | 0.50 | | .88 | | 0·65 0·98 | | 0.33 | DY87/6 | | ECL85 | 0.54 | HL42DI | |
| 5Z4G | 0.33 | 6EW6 | 0.55 | | .58 | | 0.50 | | 0.83 | DY802 | 0.29 | ECL86 | 0.33 | | 0.50 |
| 5Z4GT | | 6E5 6F1 | 0·55 0·59 | | .32 | | 0.76 | 301 | 1.00 | | 1.65 | EF22 EF40 | 0.63 | | 1·40 0·53 |
| 6/30L2 | | oF6G | 0.35 | | -88 | | 0.89 | 302 | 0.83 | E80F | 1.20 | EF41 | 0·49 0·58 | HVR2A | 0 00 |
| 6A8G | 0.33 | 6F12 | 0.17 | | .28 | | 1.00 | | 0.75 | E83F | 1.20 | EF42 | 0.33 | | 0.58 |
| 6AU7 | 0.15 | 6F13 | 0.33 | | -65 | | 0.29 | | 0.83 | | 0.60 | EF73 | 0.75 | | 0.38 |
| 6AU5 6AH6 | 0.25 | 6F14 | 0.40 | | -25 | | 0.20 | | 0.59 | | 0.40 | EF80 | 0.21 | | 0 25 |
| GAJ5 | 0.75 | 6F15 | 0.65 | | 50 | | 0.38 0.43 | | 0·10 0·53 | E180F E182CC | 0.90 | EF83 | 0.54 | | 1.75 |
| 6AJS | 0.25 | 6F18 | 0.45 | | -50 | | 0.28 | | 0.80 | | 0.53 | EF85 | 0.25 | | 0.98 |
| 6AK5 | 0.25 | 6F23 6F24 | 0.65 | | -78 | | 0.40 | | 0.50 | EA50 | 0.27 | EF86 EF89 | 0.27 | | 1·00 0·35 |
| GAK6 | 0.30 | 6F25 | 0.51 | 10C2 Q | .49 | | 0.43 | | 0.30 | EA76 | 0.88 | EF91 | 0.17 | | 0.80 |
| 6AK8 | 0.29 | 6F26 | 0.25 | | -29 | | 0.44 | 7193 | 0.53 | EABC8 | | EF92 | 0.28 | | 0.63 |
| 6AL5 | 0.10 | 6F28 | 0.60 | | -50 | | 0.26 | 7475 | 0.70 | | 0.29 | EF94 | 0.19 | KT76 | 0.63 |
| 6AM8A BANS | 0.49 | 6F32 | 0.15 | | ·50 ·75 | | 0.55 | | 1.00 | EAC91 | | EF97 | 0.55 | | 2∙00 |
| 6AQ5 | 0.21 | 6G6G | 0.25 | | 45 | | 0.74 | | 0.98 | EAF42 EAF801 | | EF98 | 0.65 | KTW61 | -63 |
| 6AQ8 | 0.32 | 6GH8A | | | .35 | | 0·69 0·61 | | 0.75 | EB34 | 0.20 | EF183 | 0.25 | KTW62 KTW63 | -63 -50 |
| 6AR5 | 0.30 | 6GK5 6GU7 | 0·50 0·50 | | .33 | | 0.58 | AC2PE | 0.98 | EB91 | 0.10 | EF184 EFξ04 | 0·27 1·20 | | 0.30 |
| 6AR6 | 1.00 | | | 10LD110 | -53 | | 0.60 | AC2PE1 | | EBC41 | 0.48 | EFP60 | 0.50 | | 0.26 |
| 6A87 | 1.00 | | 0.29 | 10LD12 | .30 | 30FL12 | | I | 0.98 | EBC81 | | EH90 | 0.34 | | 0.26 |
| 6AT6 | 0.18 | 6J6 | 0.18 | 10PL12 0 |)·30)·54 | 30FL13 | 0.47 | AC6/PE | | | 0.18 | EK90 | 0.20 | LZ339 | 0-55 |
| 6AU6 | 0.19 | 6J7G | 0.24 | | -08 | 30FL14 | | | 0.38 | | 0.28 | EL32 | 0.18 | | 0.63 |
| 6AV6 | 0.28 | 6J7(M) | 0.38 | | -28 | | 0.27 | AC/PEN | | EBF80 | | EL34 | 0.48 | | 0.75 |
| 6AW8A 6AX4 | 0.39 | 6JUSA | 0.50 | | -63 | | 0.55 | | 0.98 | EBF83 | | EL35 | 1.00 | MHLD6 | |
| 6B8G | 0.13 | 6K7G 6K8G | 0·12 0·33 | | .40 | 30L17 30P4MR | 0 65 | AC/THI | | EBF89 EBL21 | | EL41 EL81 | 0·53 0·50 | MKT4 MU12/1- | 0.98 |
| 6BA6 | 0.19 | 6L1 | 0.98 | 12AD6 0 | | | | AL60 | 0.78 | EC53 | | ELSI | 0.38 | | • 0∙38 |
| | 2 20 | | , | | | | | | | | | | 2 00 | | |

| 1 | EL84 0.21 | N308 0.95 | PM84 0.81 | UY85 0.23 | 2N966 0.58 AF178 0.75 FSY41A .25 OC23 0.42 |
|----------|--------------------------|--------------------------|--------------------------|---------------------------|---|
| | EL85 0.40 | N339 0.44 | PY33/2 .50 | U10 0.45 | 2N1756 0.55 AF180 0.58 GD4 0.86 OC24 0.42 |
| | EL86 0.38 | N379 0.28 | PY80 0.33 | U12/14 ·38 | 2N2147 0.94 AF186 0.61 GD5 0.31 OC25 0.42 |
| | EL91 0.23 | N709 0.21 | PY81 0.24 | U16 0.75 | 2N2297 0.25 AF239 0.42 GD6 0.31 OC28 0.66 |
| ı | EL95 0.32 | P61 0.40 | PY82 0.23 PY83 0.26 | U17 0.85 | 2N2369A 24 ASY27 0 47 GD8 0 22 OC29 0 69 |
| | ELL80 0.49 ELL80 0.75 | PABC80 0-32 | PY83 0.26 PY88 0.31 | U18/20 0.75 U19 1.73 | 2N2613 0.48 ASY28 0.36 CD9 0.22 OC35 0.35 |
| | EM80 0.37 | PC86 0.44 | PY301 0.65 | U22 0.39 | 2N3053 0-36 ASY29 0-55 GD11 0-22 OC36 0-47 |
| | EM81 0.37 | PC88 0.44 | PY800 0.31 | U25 0.65 | 2N3121 2·75 BA102 0·50 GD12 0·22 0C38 0·47 2N3703 0·21 BA115 0·15 GD14 0·55 0C41 0·55 |
| | EM83 0.75 | PC95 0.53 | PY801 0.31 | U26 0.60 | 2N3703 0.21 BA115 0.15 GD14 0.55 OC41 0.55 2N3709 0.22 BA116 0.28 GD15 0.44 OC42 0.69 |
| _ : | EM84 0.31 | PC97 0.36 | PZ30 0.48 | U31 0.30 | 2N3866 1 10 BA129 0 14 GD16 0 22 OC43 1 30 |
| ָ מ | EM85 1.00 | PC900 0-29 | QP21 0.50 | U33 1.50 | 2N3988 0.55 BA130 0.11 GET113 -22 OC44 0.11 |
| | EM87 0.49 | PCC84 0.27 | QQV03/10 | U35 0.83 | 28323 0.55 BA153 0.17 GET116 44 OC45 0.12 |
| 4 | EY51 0.35 | PCC85 0.24 PCC88 0.39 | 1.20 | U37 1.75 | AA119 0.17 BCY10 0.50 GET118 .22 OC46 0.17 |
| ō | EY81 0.35 | PCC88 0·39 PCC89 0·42 | QS75/20 0-63 | U45 0.78 U47 0.65 | AA120 0.17 BCY12 0.55 GET119 27 OC65 1.24 |
| Ó | EY83 0.54 | PCC189 0 46 | QS95/10 ·49 | U49 0.60 | AA129 0.17 BCY33 0.22 GET573 42 OC70 0.14 AAZ13 0.20 BCY34 0.25 GET587 42 OC71 0.12 |
| 5 | EY84 0.50 | PCF80 0.26 | QS150/15 | U50 0.30 | |
| 8 | EY87/6 0.27 EY88 0.40 | PCF82 0.30 | 0.68 | U76 0.24 | AC107 0-17 BCY38 0-25 GET8721-05 OC72 0-12 AC113 0-28 BCY39 0-28 GET873 -17 OC74 0-25 |
| b n | EY91 0.58 | PCF84 0.40 | QV04/7 0.63 | U78 0.20 | AC114 0.44 BC107 0.14 GET882 55 OC75 0.12 |
| ĭ | EZ35 0.25 | PCF86 0.44 | R11 0.98 | U153 0.24 | AC126 0.14 BC108 0.14 GET887 .25 OC76 0.17 |
| B | EZ40 0.40 | PCF87 0.74 | R16 1.75 R17 0.88 | U191 0.65 | AC127 0.19 BC113 0.28 GET889 .25 OC77 0.30 |
| 2 | EZ41 0.42 | PCF2000.67 PCF8000.55 | R18 0.50 | U192 0.23 U193 0.31 | AC128 0.22 BC115 0.17 GET890 .25 OC78 0.17 |
| 0 | EZ80 0-19 | PCF801 0 28 | R19 0.28 | U251 0.62 | AC132 0.22 BC116 0.28 GET896 .25 OC78D 0.17 AC154 0.28 BC118 0.25 GET897 .25 OC79 0.44 |
| 5 | EZ81 0-20 EZ90 0-20 | PCF8020-37 | R20 0.60 | U281 0.40 | AC154 0.28 BC118 0.25 GET897 .25 OC79 0.44 AC156 0.22 BCZ11 0.42 GET898 .25 OC81 0.12 |
| 8 | FW4/500 | PCF805 0-69 | R52 0.33 | U282 0.40 | AC157 0.28 BF154 0.28 GEX13 0.20 OC81D 0.12 |
| ก | 0.75 | PCF8060-55 | RK34 0 38 | U291 0.50 | AC165 0.28 BF158 0.32 GEX 35 0.25 OC82 0.12 |
| ž | FW4/800 | PCF8080-66 PCH200 -62 | SP13C 0.63 | U301 0-40 U329 0-62 | AC166 0.28 BF159 0.28 GEX36 0.55 OC82D 0.12 |
| 5 | 0.75 | PCL82 0.29 | SP61 0.75 TH4B 0.50 | U381 0.82 | AC167 0.66 BF163 0.22 GEX45 0.86 OC83 0.22 AC168 0.42 BF173 0.42 GEX55 0.83 OC84 0.26 |
| 4 | GY501 0.75 | PCL83 0.54 | TH233 0.98 | U403 0.33 | AC168 0.42 BF173 0.42 GEX55 0.83 OC84 0.26 AC169 0.36 BF180 0.33 GT3 0.28 OC123 0.25 |
| | GZ30 0-33 GZ32 0-39 | PCL84 0.32 | TP2620 ·98 | U404 0.38 | AC176 0.61 BF181 0.44 M1 0.17 OC139 0.25 |
| 2 | GZ33 0.70 | PCL86 0.36 | UABC80 .30 | U801 0.76 | AC177 0-31 BF185 0-44 MAT100 -43 OC140 1-05 |
| ŏ | GZ34 0.47 | PCL88 0.75 PCL800 .75 | UAF42 0.49 | U4020 0.55 | ACY17 0.28 BF194 0.17 MAT101 47 OC169 0.25 |
| 7 | GZ37 0.67 | PCL801 0 57 | UBC41 0.45 | VP2 0.53 VP13C 0.35 | ACY18 0-22 BFY50 0-25 MAT120 48 OC172 0-89 |
| 5 | HABC80 | PCL805/ | UBC81 0.40 UBF80 0.33 | VP23 0.40 | ACY19 0·21 BFY51 0·21 OA5 0·31 OC200 0·24 ACY20 0·20 BFY52 0·22 OA9 0·14 OC201 0·42 |
| | 0·44 HL13C 0·20 | PCL85 0 37 | UBF89 0.33 | VP41 0.38 | ACY21 0-21 BTX34/400 OA10 0-47 OC202 0-47 |
| 2 | HL23DD | PD500 1'44 | UBL21 0.55 | VT61A 0.35 | ACY22 0.17 2.20 OA47 0.11 OC203 0.33 |
| 8 | 0.40 | PEN4DD | UC92 0.35 | VT501 0.15 | ACY28 0-20 BY100 0-20 OA70 0-17 OC204 0-33 |
| ē | HL41DD | 1·38 PEN45 0·40 | UCC84 0.33 | VU111 0:44 VU120 0:60 | AD140 0.40 BY101 0.17 OA73 0.17 OC205 0.47 |
| 4 | 0.98 | PEN45DD | UCC85 0.33 | VU120A -60 | AD149 0.55 BY105 0.20 0A79 0.10 0C206 0.55 AD161 0.50 BY114 0.20 0A81 0.10 0C812 0.44 |
| 4 | HL42DD 0.50 | 0.75 | UCF80 0.31 | VU133 0.35 | |
| o Q | HN309 1.40 | PEN46 0.20 | UCH21 0.60 | W76 0.34 | ADDITION OF COLUMN |
| ğ | HVR2 0.53 | PEN453DD | UCH42 0.57 UCH81 0.29 | W81M 0.68 | AF102 0:99 BV V23 1:10 0 A90 0:34 50 HI 0 20 |
| 8 | HVR2A | 0.98 PENA40.98 | UCL82 0.80 | W107 0.50 W729 0.60 | AF106 0.55 BYZ10 0.28 OA91 0.10 SM1036 0.55 |
| 3 | 0.58 | PENDD | UCL83 0.54 | XE3 5.00 | AF114 0.28 BYZ11 0.28 0A95 0.10 ST1276 0.55 |
| 5 | IW3 0.38 | /4020 0·88 | UF41 0.50 | XFY12 0.48 | AF115 0.17 BYZ12 0.28 OA200 0.10 SX1/6 0.20 |
| 1 | KT2 0.25 KT8 1.75 | PFL2000.50 | UF42 0.60 | XH1-5 0.48 | AF117 0.21 BYZ13 0.28 0A202 0.11 U14706 0.28 AF121 0.38 BYZ15 1.02 0.4010 0.50 YZ20 0.98 |
| K | KT41 0.98 | PL33 0.38 | UF80 0.35 | X41 0.50 | AE124 0.98 D 1210 1.89 O A210 0.99 X 230 0.89 |
| 7 | KT44 1.00 | PL36 0.46 | UF85 0.34 | X61 0.50 | AP195 0.10 CG12E 0.22 OA211 0.75 Y543 0.20 |
| 3 | KT63 0.35 | PL81 0.42 PL81A 0.48 | UF86 0.63 | X65 0.50 X66 0.50 | AF126 0.20 CG64H 0.22 OC19 1.38 Y728 0.20 |
| 7 | KT66 0.80 | PL82 0.28 | UF89 0.27 UL41 0.54 | Z329 0.61 | AF139 0.72 FSY11A .25 OC22 0.42 ZE12V7 .10 |
| ĕ | KT74 0.63 KT76 0.63 | PL83 0.30 | UL84 0.28 | Z729 0.27 | MARGITED MD A NOVEMOD ST MS |
| 5 | KT81 2.00 | PL84 0.28 | UM80 0.33 | Z749 0.65 | MATCHED TRANSISTOR SETS LP15 (AC113, AC154, AC157, AA120). 58p per pack. |
| 5 | KTW61 63 | PL302 0.65 | UR1C 0.55 | Transistors | 1—OC81D and 2—OC81, 47p, |
| 5 | KTW62 ·63 | PL504/ | UU5 0.38 | and Diodes | 1-0C44 and 2-0C45, 47p. |
| 7 | KTW63 .50 | PL500 0.60 PL508 0.90 | UU9 0.40 | 1N1124A 58 1N4952 0 55 | 1-OC82D and 2-OC82, 53p. Set of 3-OC83 72p. |
| Ŏ | LN119 0.30 | | UU12 0.20 UY41 0.38 | | 1 watt Zeners. 2.4v., 2.7v., 3v., 3.6v., 4.3v., 4.7v., |
| 4 | LZ319 0.26 LZ329 0.26 | 12001 008 | 10121 000 | 211101 0 80 | 5 1v., 13v., 15v., 16v., 18v., 20v., 24v., 30v., 20p each. |
| * | 22020 0.20 | | | | 4 / 1 / 277 |

All goods are unused and subject to the manufacturers' guarantee. We do no handle manufacturers' seconds nor rejects, which are often described as "new and tested" but have a limited and unreliable life. Business hours Mon.-Frl. 9-5.30 p.m. Closed 1-2 p.m. Terms of business. Cash or cheque with order. Despatch charge 9p per order up to three items, then each additional item 3p extra. Orders over 65 sent free. All orders cleared same day by first class mail. Any parcel insured against damage in transit for 3p extra per order. Conditions of sale available on request. S.A.E. with all enquiries, please. Regret no lists this month.

Samson's

9 & 10 CHAPEL ST., LONDON, N.W.I 01-262-5125 01-723-7851

| | FULLY | SHROUDED | (excepted | L.T. TRANS- D) TERMINAL |
|----------|---------|-------------|-----------|----------------------------|
| BLOCK CO | NNECTIO | NS, ALL PRI | MARIES | 220/240v. |

| Type No. | Sec. Vol | t Taps. | | | | Amps. | Price | Carr. |
|----------|-----------|---------|-----|-----|-----|-------|--------|-------|
| 1A | 25-33-40- | 50 | | | | 15 | £13·00 | 75p |
| 1B | 25-33-40- | 50 | | | | 10 | £11.00 | 50p |
| 1C | 25-33-40- | 50 | | | | 6 | £7·75 | 50o |
| 1D | 25-33-40- | 50 | | | | 3 | £5 75 | 50p |
| 2Ă | 4-16-24-3 | 32 | | | | 12 | £8·75 | 50p |
| 2B | 4-16-24-3 | | | | | 8 | £7.60 | 50p |
| žČ | 4-16-24-3 | | | | | 4 - | £4·75 | 40p |
| 2D | 4-16-24- | | | | | ż | £3.20 | 30p |
| 3Ã | 24-30-36 | | | | | 10 | £8·25 | 60p |
| 3B | 24-30-36 | | | | | 5 | £6.25 | 40p |
| зС | 24-30-36 | | | | | ž | £3.25 | 30p |
| 4A | 12-20-24 | | | | | 20 | £10.95 | 75p |
| 4B | 12-20-24 | | | | | 10 | £6.50 | 40p |
| 4C | 12-20-24 | | | | | Š | £4:35 | 30p |
| 5A | 3-12-18 | | • • | | | 20 | €9.76 | 50p |
| 5B | 3-12-18 | | • • | | :: | 10 | £6.00 | 45p |
| 5C | 3-12-18 | •• | • • | • • | | 5 | £4 10 | 35p |
| 6A | 48-56-60 | :: | • • | • • | | 2 | £4-10 | 35p |
| 6B | 48-56-60 | | • • | •• | | ī | £3.00 | 30p |
| 7A | 6-12 | •• | • • | • • | • • | 20 | £7.50 | 50p |
| | | • • | • • | • • | • • | 10 | £4·10 | 40p |
| 7B | 6-12 | | • • | • • | • • | 5 | £2.75 | 40p |
| 7C | 6-12 | • • | • • | • • | • • | | £2·10 | 30p |
| 8A . | 12-24 | • • • | • • | • • | • • | 1 | | 25p |
| .9A | 17-32 | • • | | • • | • • | 8 | £7.00 | 45p |
| 10A | 9-15 | • • | | • • | • • | .2 | £2·10 | 25p |
| 11 A | - 6⋅3 | | | | | 15 | £4·10 | 30p |

Note: By using the intermediate taps many other voltages can be obtained.

Example: No. 1 ... 7-8-10-15-17-25-33-40-50v.
No. 2 ... 48-12-16-20-24-32v.
No. 5 ... 3-6-9-12-15-18v.

30-25-0-25-30v. 2a. with Screen £4-30, carr. 35p. 36-0-36v. 5a. £9-40 carr. 50p. As recommended for Linsley Hood Amplifiers.

| HEAVY LEADS | DUTY L | | | TY | PE 9 I | NCH F | YING |
|----------------|-----------|------|------|----|--------|--------|-------|
| Type No. | Sec. Volt | Tap. | | | Amps. | Price | Carr. |
| 1 | 24-30-36 | | | | 20 | £14·00 | 75p |
| 2 | 12-20-24 | | | | 30 | £13 00 | 75p |
| 3 | 3-12-18 | | | | 30 . | £13·00 | 75p |
| 4 | 6-12 | | | | 50 | £14.00 | 75p |
| 5 | 20 | | | | 30 | £12·00 | 75p |

DAVENSET ISOLATION TRANSFORMERS
Pri. 10-0-200-220-240v, Sec. 240v. Centre tapped 1-2kva.
Conservatively rated. Size 8½ x 7 x 8½ ins. Wgt, 59 lbs. Open frame type, terminal connections. Fraction of maker's price.
£17-80 carr. £1-00.

H.T. TRANSFORMERS BY FAMOUS MANUFACTURERS
PARMEKO. All primaries 220-240v. Type 1. Sec. 630-0-620v. 105mla 5v. 4A. 5v. 2A. Potted type £3-60. Carr. 50p. Type 2. Sec. 1,875v. 60mla. 42kv. wks, and 500v. 31mla. Potted type £3-60. Carr. 50p. Type 3. Sec. 310-0-310v. 35mla and 200-0-200v. 20mla. 6-3v. 1A. 6-3v. 1A. Potted type £2-75. P.P. 50p. Type 4. Sec. tapped 760-700v. 50mla. 6-3v. 1-5a. £1-75. P.P. 30p.

WODEN. All primaries 220-240v. Type 1. Sec. 830-710-0-710-880v. 120m/a. unstrouded table top connections, tropicalised £2:50. P.P. 50p. Type 2. Sec. 190v. 60m/a. 6:3v. 3a. £1:25. P.P. 25p. Type 3. Sec. tapped 150-165v. 4 amps unstrouded table top connections £3:75. P.P. 75p. Type 4. Sec. 130v. 450m/a. three times. "C" core, table top connections £3:50. P.P. 50p. Type 5. 63v. 1-6a. and 24v. 0-8a. and 6:3v. 1a. unshrouded table top connections £2:50. Carr. 50p.

GARDNERS. All primaries 220-240v. Type 1, 350-0-350v. 60m/a. 6·3v. 4a. 5v. 2·5a. shrouded £1·50. P.P. 30p. Type 2, 300-0-300v. 60m/a. 6·3v. 4a. "C" core. £1·50. P.P. 30p. Type 3, 450-400-350-0-350-400-450v. 50m/a. "C" core £1·25. P.P. 25p. Type 4. 250-0-250v. 100m/a. 6·3v. 3a. 6·3v. 3a. 5v. 3a. Potted type £2·50. P.P. 55p. Type 5. 350v. 44m/ja. 20v. 10m/a. 6·3v. 3a. "C" core £1·50. P.P. 30p.

6.3v. 3a. "C" core £1.5v. P.P. 30p.

L.T. TRANSFORMERS

WODEN Pri. 220-230-240-250v. Sec. 25v.
2a. Twice, 16v. 4a. twice, 26v. 4a. 31v.
7a. All separate windings. Conservatively rated, Open frame type table top connections. Size 6½ x 6 x 6in. £8.50, carr. 50p.

AMOS 'C' core type. Pri. 200-220-240v.
Sec. 180-18v. 5a., and 18-0-18v. 3a. Conservatively rated table top connections. \$2.50 P.P. 50p.

REDCLIFFE 'C' core types. Pri. 220-240v.
Sec. 35v. 350 m/la. 75p. P.P. 25p. Pri. 220-240v.
Sec. 36v. 350 m/la. 75p. P.P. 25p. Pri. 220-240v.
Sec. tapped 370-390-400v. 6 m/la. 50p.
P.P. 20p.

P.P. 20p.

G.E.C. L.T. TRANSFORMERS

All Primaries 220-240v. Type 1 tapped.
63-68-74v, 3a. and 6-3v. 4a. terminal block
connections. Unshrouded £3-00. P.P. 50p.
Type 2 tapped. 59-61-65-67-69v. 10a. T
block connections. Unshrouded £6-00
carr. 75p. Type 3 tapped. 56-58-60v. 3a.
T block connections. £2-95. P.P. 50p.
Type 4 100-0-100v. 65m/a. and 61-64-67v.
T50m/a. and 6v. 1a. £1-75. P.P. 25p.
Type 5 tapped. 11-5-13-5v. 14a. and 13-5v.
1-4a. twice. Unshrouded, T block connections.
£4-50 carr. 75p. Type 6 10v. 2a. and
50v. 0-6a. T block connections. Unshrouded. £1-50 carr. 25p. Type 7 15v. 4a.
and 13v. 6a. T block connections. Unshrouded. £3-carr. 50p. Type 8 17v. 2a.
twice, unshrouded. £1-75. P.P. 30p.

AMOS 'C' CORE TRANSFORMERS Pri. 220-240v. Sec. 53-6v. 6a. £3-75. P.P. 50p. Pri. 240v. 17-5v. 6a. £2-25. P.P. 35p.

CRESHAM POTTED OIL-FILLED CHOKES

9 henries 500m/a. 5kv. wkg. £7:50 carr. £1:50. AMOS L.T. CHOKES core type. 140m/h. 5 amps. £4·50 carr.

G.E.C. LT TRANSFORMERS
Pri. 200-220-240v. Three separate
Secs. 27v. 9A., 9v. 9A., 3v. 9A.
The following voltages can be
obtained: 3-9-12-27-30-36-39v. 9A.
Open frame. Fully tropicalised.
Table top connections. £4-50 carr.
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H.T. TRANSFORMERS PARMEKO. Pri, 240v. Sec. 250-0-250v. 50 m/a. 6:3v. 1a. £1'25. P.P. 35p., size 4 × 3 × 2½ ins. GARDNERS. C. core. Pri. 240v. Sec. 300-0-300v. 66 m/a. 6:3v. 4a. £1:50. P.P. 35p., A.C.I. Pri. 240v. Sec. 250v. 60 m/a. 15v. 12a. 6:3v. 4:5a. £1:25. P.P. 35p., open type table top connections. Size 4 × 3½ × 3 ins.

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| 64 | 75 | 1 14 | 7.0 x 6.4 x 6.0 | 0-115-210-240 | i ·82 | 30 |
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| 66 | 300 | 6 0 | 10.2 × 10.2 × 9.5 | | 4.28 | 52 |
| 67 | 500 | 12 š | 14.0 × 10.2 × 11.4 | " " | 6.35 | 67 |
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| 111 213 | 0.5 0.25 1.0 0.5 | I 0 | 7.6 × 5.7 × 4.4 8.3 × 5.1 × 5.1 | 0-12V at 0-25A × 2 0-12V at 0-5A × 2 | 0.93 22 1.11 22 |
| 777 | 2 1 | iŏ | 7.0 × 6.4 × 5.7 | 0-12V at 1A X 2 | 1.46 22 |
| 18 | | | 8·3 × 7·0 × 7·0 | 0-12V at 2A X 2 | 2.04 36 |
| . 70 | 4 2 6 3 8 4 10 5 16 8 | 2 4 3 12 5 4 6 3 | 10.2 × 7.6 × 8.6 | 0-12V at 3A x 2 | 2.46 42 2.73 52 3.23 52 |
| 108 72 | 10 5 | 6 3 | 10.0 × 8.3 × 8.2 7.9 × 10.8 × 10.2 | 0-12V at 4A × 2 0-12V at 5A × 2 | 3·23 52 |
| iĩ | i6 8 | 78 | 12·1 × 9·5 × 10·2 | 0-12V at 8A x2 | 4.99 52 |
| 115 | 20 10 | 11 13 | 12·1 × 11·4 × 10·2 | 0-12V at 10A x 2 | 6.35 67 |
| 187 226 | 30 I5 60 30 | 16 12 | 13·3 × 12·1 × 12·1 17·0 × 14·5 × 12·5 | 0-12V at 15A × 2 | 11.73 82 21.57 * |
| 226 | 60 30 | 34 0 | 17.0 X 14.5 X 12.5 | 0-12V at 30A ×2 | 71.21 |
| Ref. | Amos. | Weight | Size cm. | 30 VOLT RANGE Secondary Taps | P & P |
| No. | Amps. | ib oz. | Size Cill. | secondary raps | |
| 112 | 0.5 | 1 4 | 8.3 x 3.7 x 4.9 | 0-12-15-20-2#-30V | 1-11 22 |
| 79 | 1.0 | 2 0 | 7.0 x 6.4 x 6.0 | ** ** | 1.48 36 |
| 3 20 | 2·0 3·0 | 2 0 3 2 4 6 6 0 6 8 7 8 | 8.9 × 7.0 × 7.6 10.2 × 8.9 × 8.6 | | 2·21 36 2·72 42 |
| 21 | 4.0 | 6 0 | 10.2 × 10.0 × 8.6 | 17 17 | 3.23 52 |
| 51 | 5∙0 | 6 0 6 8 7 8 | 12·1 x 10·0 x 8·6 | ., ,, | 4.02 52 |
| 117 | 6.0 | 78 100 | 12·1 × 10·0 × 10·2 | " " | 4.80 52 |
| 88 89 | 8·0 10·0 | 10 0 12 2 | 14.0 × 11.7 × 10.0 14.0 × 10.2 × 11.4 | ., ., | 6·20 67 7·85 67 |
| 67 | 10.0 | 12 2 | 14.0 × 10.2 × 11.4 | 50 VOLT RANGE | 7.03 |
| Ref. | Amps. | Weight | Size cm. | Secondary Taps | P & P |
| No. | - | lb oz | | | £ p |
| 102 | 0.5 | 1 11 | 7.0 × 7.0 × 5.7 | 0-19-25-33-40-50V | 1.46 30 |
| 103 | 1.0 2.0 | 2 10 5 0 | 8·3 x 7·3 x 7·0 10·2 x 8·9 x 8·6 | 11 15 | 2·13 36 2·96 42 |
| 105 | 3.0 | 6 0 | 10.2 × 10.2 × 8.3 | " " | 4.01 52 |
| 106 | 4.0 | 9 4 | 12·1 x 11·4 x 10·2 | " " | 5.31 52 |
| 107 | 6.0 | 12 4 | $12 \cdot 1 \times 11 \cdot 1 \times 13 \cdot 3$ | ,, ,, | 7.85 67 |
| 118 | 0.01 | 18 9 | 13·3 × 13·3 × 12·1 16·5 × 11·4 × 15·9 | " " | 10·25 97 12·85 97 |
| | | | Size cm. | 60 VOLT RANGE | P & P |
| Ref. | Amps. | Weight Ib oz | size cm. | 80 VOLI RANGE | £ ' \$ |
| 124 | 0.5 | 2 4 | 8.3 × 9.5 × 6.7 | 0-24-30-40-48-60V | 1.48 36 |
| 126 | 1.0 | 3 0 | 8.9 x 7.6 x 7.6 | ,, ,, | 2.06 36 |
| 127 | 2·0 3·0 | 5 6 8 8 | 10.2 x 8.9 x 8.6 11.9 x 9.5 x 10.0 | 11 11 | 3·23 42 4·92 52 |
| 123 | 4.0 | 10 6 | 11.4× 9.5×11.4 | 11 17 | 6.35 67 |
| 120 | 6·0 | 16 12 | 13·3 x 2·1 x 2·1 | 11 17 | 9.20 82 |
| 122 | 10.0 | 23 2 | 16·5 × 12·7 × 16·5 | ,, ,, | 15.23 * |
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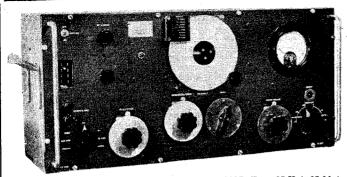
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T.1509 TRANSMITTERS (FOR EXPORT ONLY): General-purpose HF communications transmitter for use in fixed or mobile ground stations. Hand or high-speed keying. Crystal or MO control, with temperature compensated MO circuit.CW, MCW and R/T. Frequency: 1.5 to 20 Mc/s. Modulation: 100% O/put impedance: 50 ohms. Audio input: 600 ohms. Valves: Power Amplifier 2 × 813 and Modulator 2 × 813. Power requirements 200-250 volts a.c., 50 cycles. Power out put 300 watts. Dimensions 2ft. 6in. W. × 2ft. D. × 5ft. H. Weight: 800 lbs. Excellent condition, price 225.00 each. AN/ARC-27 TRANSMITTER/RECEIVER (FOR EXPORT ONLY). Frequency 225-400 mc. 1750 channels 100 Kc apart with 18 preset channels Modulation: am. Power output 9 watts. Receiver is superheterodyne. Max. output 2 watts. Antenna: 50 ohm impedance. Power requirements 24v d.c. Complete transmitter with operating cables, control box, headphones, microphone. Price £250.00 each secondhand, excellent condition.

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TS-683 CRYSTAL IMPEDANCE METER: 10-140MHz secondhand good condition. £75·00 each +£1·00 carr.

TUNING UNIT: 24V geared motor driving double 25pf double spaced variable capacitor. One m/c relay and 2 other relays. £2·50 each 30p post, good condition. UHF ASSEMBLY: (suitable for 1,000MHz conversion) including UHF valves: 2C42, 2C46, 1840 (complete with associated capacitors and screening), 3 manual counters 0-999. Valves 6AL5 and 8×6AK5. £10·00 plus 60p post, good condition. MODULATOR UNIT: complete with transformer and 2×807 valves mounted in 19 in. chassis \times 8 in. high \times 8 in. deep. £4.50 secondhand cond., or £6.56 new cond. Carriage £1.

RF UNIT: suitable for use with the above unit. Complete with $2\times3E29$ valves. Ideal for conversion to 4 metres. £5 secondhand cond., or £7.50 new cond. Carriage £1.

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| 18/11 18/11 A | 25µf 24µf | 12v 275v | 4p 4p | H6/1 A H6/2 | 250µf 250µf | 4v 25v | 3p 3p | 0/2 101/2 | 00 | 1000 + 1000 | r o amps | 302 | 0.1 |
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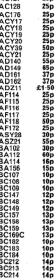


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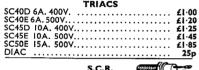












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| CRS 7/60 7A. 600V. 2N4444 8A. 600V. | · · | • • | • • | • • | | | • | • • | : | : | • | • | | 98 £1: |
| BRIDG | E | RI | ĒC | T | TF | 1 | Ę | R | s | | | _ | _ | |

| BRIDGE | | | | |
|--------------------|----|---|-----|---|
| BY122 1.5A. 40V | | | | |
| BY164 1.4A. 200V | | | | Ĩ |
| CIC2-100 2A. 100V | ٠. | • | • • | • |
| J. J. 100 Z. 100 1 | | • | • • | • |

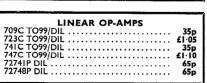
| | ZENER DIC | DES |
|------|-----------|--------|
| 7000 | | 2 221/ |

| 10 Wattrange | 45p |
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| BZY88 series 400mW. 3.3-33V. 5% 1.5 Watt range | 15p 25p |

TIL 209 H.P. 5082

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ÓRP 12



MOVING IRON AMMETERS 2년 in. SQUARE

Available in the following values:-

Available in the 0-1-4 Amp 0-1-5-6 Amp 0-1-5-9 Amp 0-5-30 Amp 0-8-48 Amp 0-15-45 Amp 0-30-180 Amp 0-40-240 Amp 0-50-200 Amp 0-50-300 Amp

All Brand New and Boxed ONLY £1.75

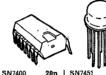
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50p

| Ī | PA230 | £1-10 | l Watt Audio Amp. |
|---|----------|-------|------------------------|
| l | PA234 | £1.25 | 2/3 Watt Audio Amp. |
| l | PA246 | £1.75 | 5 Watt Audio Amp. |
| l | CA3014 | £1-55 | |
| l | CA3018 | £1.00 | * |
| ١ | CA3048 | £2.34 | Stereo Pre-Amp. |
| ı | MC1303L | £1.85 | Stereo Pre-Amp. |
| | MFC4000 | 55p | 250mWatt Audio I.C. |
| | MFC4000A | 60p | |
| | SL403D | £1·50 | 3 Watt Audio Amp. |
| | ZN414 | £1·25 | Radio I.C. |
| | I M309K | 41.00 | 5V IA Voltage Reg I.C. |

DIGITAL







| ▼. | ₩ U: | , | - 4 | |
|--------------------------|--------------------------|--------------|--------------------|----------------|
| SN7400 201 | I SN7451 | 20p | SN74150 | £3·35 |
| SN7401 20 | SN7453 | 20p | SN74151 | £1 10 |
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| SN7408 201 | SN7475 | 55p | SN74160 | £2.60 |
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| SN7433 70 | | £2.50 | SN74184 | £2·45 |
| SN7437 65 | SN74104 | £1·45 | SN74185A | £2·40 |
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| SN7443 £1.0 | | £1.90 | SN74194 | £2.50 |
| SN7445 £2.0 | 0 SN74121 | 60p | SN74195 | £1·85 |

V.A.T.

Unless otherwise stated all prices are **EXCLU-SIVE** of V.A.T. Please add 10% to all orders. Carriage: orders under £5 + 20p. Over £5

COMMUNICATIONS EQUIPMENT POCKET V.H.F. F.M. RADIOTELEPHONE Cossor Type CC2/8 Mk. 2.

Fully transitorised transmitter/receiver available in two versions:-

transintreer/receiver available in two versions:—

Low band; Freq. range 71.5-104MHz.

R.F. Output 500mW.

Complete with \$\frac{1}{2}\$ wave whip aerial, combined microphone/loudspeaker and 13.3V. rechargeable nickel-cadmium DEAC battery Price £75 + v.a.t.

U.H.F. 2 watt FIXED RADIO LINK.

2V. dc/240V. ac. F.M. TRANSMITTER/Type CC RTX 4A Mk. 1

R.F. Output 2W at 450-470MHz.

RECEIVER/Type CC RR4A Mk. 1

Price £80.00 per unit Full Technical and operating data available. Prices and details on request. Mains Power Pack for the above

I + I CARRIER EQUIPMENTS. Cossor Type CC M2A.

Solid state multiplex installations designed for U.H.F. radio systems enabling 2 speech channels each with out of band signalling, if required or the equivalent in telemetry information, to be transmitted simultaneously over a radio system.

Prices and details on request.

V.H.F. RADIOTELEPHONE BASE STATION. Cossor Type CC 603 Transmitter. Simplex or duplex operation, local or remote control with talk through facilities, using double sideband a.m. modulation.

Low-band 71.5-104MHz. or High-band 156-174MHz.

Low-band 71.5-104MHz. or High-band 156-1/4MHz. versions available. RF. Output power 25W. into 50 Ohms. 24V. dc. operation. Prices and details on request OPTIONAL POWER SUPPLY Type CC 101 for type CC603 base station P.O.A.

SELECTIVE CALL SYSTEM. Codar Type CC 505/50 (50 way) or CC 505/100 (100 way). The Cossor selective call system may be used with any communication system where a base station is required to call any one or all of a number of sub-stations. Both versions available, all new and sub-stations. Doi: version: in original packing. Price: 50 way £65 + v.a.t. 100 way £80 + v.a.t.

DEAC RECHARGEABLE BATTERY CASSETTES 13.4V (nom.) type B/SA 80351/108 Heavy duty encapsulated DEAC supply. Size $3\frac{1}{2} \times 2\frac{1}{2} \times 1\frac{1}{2}$ in. Price £5 + v.a.t.

8-WAY BATTERY CHARGER Type CC 999
Charges up to 8 of the above battery cassettes.
Price £25 + v.a.t.

I2-WAY BATTERY CHARGER Type CC 999 Charges up to 12 of 13.4V DEAC batteries. Metered battery condition check. Price £35 + v.a.t.

MICROPHONES S. G. Brown Stick Microphone and Stand. Push-to-talk button. 300Ω. £5 complete. S. G. Brown Hand-held with push-to-talk button. £8 each.

OSCILLOSCOPES

CAWKELL
Revscope S.O.I Storage Scopes..... from £100
COSSOR COSSOR
CDU.110. Double beam DC—20MHz. Brand new with manual £300
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 TEKTRONIX

 536 X-Y Oscilloscope. DC—I1MHz
 £300

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 £310

 545A DC—30MHz
 £385

 551 Dual Beam. 27MHz
 £300

 661 Sampling Scope. IGHZ 2m V/cm. Dual beam. £475
 £610

 661 Sampling Scope. 3.9 GHz. 2m V/cm. Dual beam. £475

ELECTRONIC COMPONENTS BARGAIN COMPONENT PACKS

No.

1 500 Carbon resistors, ½, ½, 1, 2 watt.
2 100 Electrolytic Condensers.
3 250 Ceramic, Polystyrene, Silver Mica, etc.,

4 250 Polyester, Polycarbonate, Paper, etc., Condensers. 5 25 Potentiometers, assorted.

250 High-stab. 1%, 2%, 5% resistors. 50 Assorted Tagstrips. 11b Assorted nuts, bolts, washers, spacers, etc. 25 Assorted switches, rotary, lever, micro,

toggle, etc. 10 50 Preset Potentiometers.

Trial mixed component pack £1.
2 Jumbo mixed pack £5.
LL COMPONENTS NEW AND UNUSED
£1 + 25p p.p. per pack, £5 for 5 packs p/free.

SEMICONDUCTORS

COMPONENTS FOR W.W. AMPLIFIER DESIGNS

| AA'AA' WIAILE | -11-1 | En | D L | ,,,,, | 10 | |
|--|---------|----------|---------|----------|-----------|--------|
| 100W AMPLIFIER (FEB. | 1972 |) | | | | |
| Designer approved kit. | | | | | | |
| Semiconductor set | | | | | | 15-60 |
| Resistors, capacitors, pots | | | | | | 2.50 |
| F/Glass PCB | | | | | | 1 - 30 |
| POWER SUPPLY (For | | Amp | .) | | | |
| Designer approved kit. | | | • | | | |
| Semiconductors, Resistors | . car | acitor | s. po | ts. tr | ans- | |
| | | | | | | 14.70 |
| 30W BLOMLEY (New as | | | | | | |
| Semiconductor set | | | | | | 5.60 |
| Resistors, capacitors, pots | | :: | | :: | | 1 . 85 |
| F/Glass PCB | | •• | | :: | | 0.70 |
| 30W BAILEY (Single pow | er rai | n' | | | | |
| Transistor set | | ·/ | | | | 4.60 |
| Transistor set | | | | | | 1 - 45 |
| F/Glass PCB | | | | | | 0.65 |
| LINSLEY-HOOD CLAS | S A (| Dec 1 | 970 c | ircuit) | | |
| Designer approved kit. | 3 ~ (| Dec., | ,, c | ii cuit) | | |
| 2N3055 pair, BC212L, 2N17 | 71: | | | | | 1 · 20 |
| Resistors, capacitors, pot | | | • • • | | | 1.80 |
| E/CI BCD | | | | • • • | | 0.60 |
| LINSLEY-HOOD 20W | CI A 6 | e A D | | • • | • • | |
| LINSLET-HOOD 20W | CLAS | 3 AD | | | | |
| | | | | | | 3 · 35 |
| MJ481/491, MJE521, BC182L | ., DCZ | ZL, Ze | itei | • • | • • | |
| Resistors, capacitors, pots | | | | • • | • • | 0.70 |
| F/Glass PCB Please state 8Ω or 15Ω | • • | • • | • • | •• | • • | 0.70 |
| | /FD 0 | | · · | | | |
| REGULATED 60V POW | | | | • | | |
| A 5 transistor series stab | iliser, | suitab | ie for | a pai | rot | |
| Bailey or Blomley amplifi | ers, 16 | aturin | g ver | у епес | tive | 4.85 |
| S/C protection. All Semi/C' | s, K's, | US, F | Glass | rcb | - i - i - | 4.03 |
| Power supplies for o | | | | | iDie | |
| BAILEY/BURROWS PR | E-AM | P (Aug | ,, 1971 |) | | |
| Component Set: Mono Component set: Stereo | • • | • • | • • | • • | • • | 2.75 |
| Component set: Stereo | | | •:- | | • • | 6 · 35 |
| Each component set compr | rises o | i ali sp | ecitiec | resist | ors, | |
| capacitors, transistors, po | ts, inc | luding | specie | al balo | ince | |
| control for stereo sets. | | | | | | |
| Stereo F/Glass PCB | | • • | • • | • • | • • | I · 60 |
| STUART TAPE RECOR | | | | | | |
| Set of stereo f/glass PCBs | | • • | | | | 2.70 |
| Components sets on price ! | ist. | | | | | |
| | | | | | | |

TEXAN' TEXAS INSTRUMENTS DESIGNED & APPROVED FULL KIT



£28.50 INCLUDES TEAK CASE

20 Watt per channel stereo amplifier designed by Richard Mann of Texas Instruments and published in Practical Wireless May-July 1972.

This low distortion (0.09% at 20W into 8 ohm), wide bandwidth (-3dB 5Hz-35KHz) design is offered as a Texas Instruments approved full kit (including all metalwork and Teak case for a total of £28.50 post paid. Full details in price list.

METALWORK SYSTEM

Designed to house Bailey, Blomley or Linsley Hood Class AB amplifiers with simple or regulated power supplies and Bailey Burrows pre-amp. Options of standard or hum reducing toroidal mains transformer.

| TOROIDAL TRANSFORMER 60 volt 2 amp. | * |
|--|-------|
| Max. height 2in. Suitable for our regulated power supply | £7.40 |
| Simple clamp | £0.20 |
| Magnetically screening clamp | £0.75 |

2N699 2N1613 0.20 BC212L 0.12 0·14 0.13 2N171 0.25 2N2926G BCY72 0.10 2N3053 2N3055 0.15 BF257 BF259 0.47 0.45 2N3702 0.11 BFR79 0.25 2N3703 2N3704 0.10 0.10 MJ481 MJ491 2N3706 0.09 2N3707 0.10 MIE521 MPSA05 MPSA12 MPSA14 MPSA55 MPSA65 MPSA65 MPSU55 SN72741P SN72748P THB11 TIP29A TIP30A TIP31A TIP31A TIP31A TIP32A TIP31A TIP31A 2N3708 0.07 0.09 2N3709 2N3710 0.09 2N3711 0·35 0·40 0·60 0·23 0·17 2N3819 2N3904 2N3906 2N4058 0·20 0·12 2N4062 2N4302 0.60 0.50 0.60 0.60 0.70 1.00 1.50 0.74 0.90 0.42 0.54 2N5087 2N5210 2N5830 0.30 40361 0·45 0·08 40362 BC107 BC108 BC109 0·08 0·08 0·15 0·15 **BC125** BC126 0.05 0.10 0.25 1.20 1544 15920 BC182K 0·10 0·12 BC212K

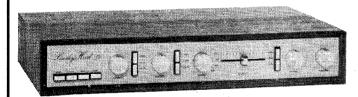
HI-FI NEWS 75 WATT AMPLIFIER

BY J. L. LINSLEY-HOOD

Published Nov. 1972 to Feb. 1973

BC182L

DESIGNER APPROVED KIT



SLIMLINE STYLE CHASSIS DIMENSIONS: 17.0in. x 2.0in. x 12.0in. This slimline unit has been made practical by the use of a specially designed TOROIDAL TRANSFORMER and highly compact printed circuit boards which have been fully tested and approved by Mr. Linsley-Hood.

FREE TEAK CASE

Total cost of individually purchased packs: £63.95

WITH 75 WATT PER CHANNEL COMPLETE AMPLIFIER KITS

Cost of complete kit:

£56,60

TRADE ENQUIRIES WELCOME

P.S. Full circuit description in handbook 30р

FOR FURTHER DETAILS PLEASE WRITE TO:

POWERTRAN ELECTRONICS

PORTWAY INDUSTRIAL ESTATE, ANDOVER : HANTS

MAIL ORDER ONLY

POST FREE TO U.K.

OVERSEAS AT COST

U.K. Orders Subject to 10% V.A.T. Surcharge

75 WATTS PER CHANNEL BANDWIDTH (3dB) 3HZ-40KHZ DISTORTION LESS THAN 0.01% UNCONDITIONAL STABILITY

| | COMPONENT PACKS | |
|-----|---|-------|
| Pac | ···· | |
| ı | Fibre glass printed circuit board for power amp | £0.75 |
| 2 | Set of resistors, capacitors, pre-sets for power amp | £1.50 |
| 3 | Set of semi-conductors for power amp. (highest voltage version) | £5.50 |
| 4 | Pair of 2 drilled, finned heat sinks | £0.80 |
| 5 | Fibre glass printed circuit board for pre-amp | £1.10 |
| 6 | Set of low noise resistors, capacitors, pre-sets for pre-amp | £2.70 |
| _ | · | £2.10 |
| 7 | Set of low noise, high gain semi-conductors for pre-amp | |
| 8 | Set of potentiometers (including mains switch) | £1.55 |
| 9 | Set of 4 push button switches, rotary mode switch | £3.10 |
| 10 | Toroidal transformer complete with magnetic screen/housing primary: 0-117-234 V. secondaries: 33-0-33 V. 24-0-24 V., electrostatic screen | £9.15 |
| П | Fibre glass printed circuit board for power supply | £0.55 |
| 12 | Set of resistors, capacitors, secondary fuses, semi- conductors for power supply | £3.50 |
| 13 | Set of miscellaneous parts including DIN skts., mains input skt. fuse holder, interconnecting cable, control knobs | £3.25 |
| 14 | fascia panel and all brackets, fixing parts, etc. | £6.30 |
| 15 | Handbook, based on Hi-Fi News articles | £0.30 |
| 16 | Teak cabinet | £7.35 |
| | Basic Component Set | |

Basic Component Set

Set of semi-conductors, resistors, capacitors, printed circuit boards for stereo power amp, pre-amp. and power supply.

£31.35

Handbook Included



Sensational 'Once in a Life-time Offer' Unrepeatable once sooks are cleared. The amplifier is made by Mullard. Carries maker's Guarantee. In nest case. May be used for Mono or Stereo. Music or speech, Works off dry battery, ear batt. or mains power pack. FREE—all who purchase we send copy of Mullard booklet DIY stereo. 21:60



DRILL CONTROLLER

New 1kW model.

Electronically changes speed from approximately 10 revs. to maximum. Full power at all speeds by inger-tip control. Kit includes all parts, case, everything and full instructions g1.65, plus 13p post and insuvance. Made up model also available £2.48 plus 13p p. & p

MIGHTY MIDGET
Probably the finest possible radio, as described in Practical
Wireless, January 73. All electronic parts £2:20 post paid.

GOOD COMPANION I.C. VERSION



We can now offer these again in I.C. version using Ferranti ZN414 and Mullard AF Module 1172. Cabinet size approx. Ilin. wide X Sin. high X Sin. deep. Complete assembly instructions, £6:33 plus 25p post and ins. Excellent tone wood cabinet.

CHIP RADIO

Ferranti's latest device ZN414—gives results better than superhet. Supplied complete with technical notes and circuits £1.38 each. 10 for £11.11.

HI-Q TUNER COMPONENTS

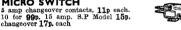
HI-Q TUNER COMPONENTS
For experimenting with the ZN414
Kit No. 1 Plessey Ministure Tuning Condenser with built in LW switch and 3in. Ferrit slab and litz wound MW coil 729.
Kit No. 2 Air spaced tuning condenser 6in. ferrit rod litz wound MW and LW coils 949.
Kit No. 3 Air spaced TO with slow motion drive 8in. ferrit rod with litz wound LW and MW coils 21.10.
Kit No. 4 Permeability tuner with fast and slow motion drive and the Soleman
Permeability tuner with fast and slow motion drive and LW loading coils. 50p.



12 VOLT 1½ AMP
POWER PÄCK
This comprises double-wound 230/
240V mains transformer with full
wave rectifier and 2000 mf/d
smoothing. Price \$2.20, plus 20p
post & packing.

Heavy Duty Mains Power Pack, Output voltage adjustable from 15-40V in steps—maximum load 250W—that is from 6 amp at 40V to 15 amp at 15V. This really is a high power heavy duty unit with dozens of workshop uses. Output voltage adjustment is very quick—simply interchange push on leads. Silicon rectifiers and smoothing by 3,000mF. Price £6-35 plus 65p post.

MICRO SWITCH





MAINS RELAY BARGAIN

Special this month are some single, double and treble pole changeover relays. Contacts rated at 15 amps. Operating coil would for 240V. A. C. Good British Mak. Unused Size approx. 1½ x lin. Open construction. Single pole 25p each 10 for £2.48 Treble pole 39p each 10 for £3.47

MAINS OPERATED SOLENOIDS



Model 772—small but powerful
lin, pull — approx. size 1½ × 1½
× 1½in., 66p.
Model 4001/—¾in. pull. Size
2½ × 2≥ × 14in. 83p.
Model TTI—1½in. pull. Size
3 × 2½ × 2in. 82p.
Model approx. Size 3 × 2½ × 2in. 82p.
Model TTI—1½in. pull. Size
3 × 2½ × 2in. 82p.

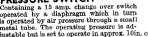
MAINS TRANSISTOR POWER PACK

PACK Designed to operate transistor ests and amplifiers. Adjust-able output 6v., 9v., 12 volts for up to 500mA (class B woring). Takes the place of any of the following batteries: PP1, PP3, PP4, PP6, PP7, PP9, and others. Kit comprises: mains transformer rectifier, smoothing and load resistor condensers and instructions. Real snip at only 47.10 plus 200 nostate. resistor condensers and £1.10 plus 20p postage.



TIME SWITCH
Smith's mains driven clock with 15
amp switch, also notes showing how
you can wake up with music playing,
kettle boiling or come home to a wanhouse, warn off burglars, keep pets
warm, halve your heating bill, etc.
\$2.25. **C**

PRESSURE SWITCH



PRESURE SWITCH
Containing a 15 amp. change over switch operated by a diaphragm which in turn is operated by all pressure through a small metal tube. The operating pressure is adjusted by the second of the second

HONEYWELL PROGRAMMER



HONEYWELL PROGRAMMER

This is a drum type timing device, the drum being calibrated in equal divisions for switch setting purposes with trips which are infinitely adjustable for position. They are also arranged to allow 2 operations per switch per rotation. There are 15 change-over micro switches each of 10 amp type operated by the trips thus 15 circuits may be changed per revolution. Drive motor is mains operated 5 revs per min. Some of the many uses of this timer sre Machinery control. Boller dring, Dispensing and Vending machines. Display lighting animated and signs, Signalling, etc. Price from makers probably over \$10 each. Special snip price \$6:38 plus 25p post and insurance. Don't miss this terrific bargain.

THYRISTOR LIGHT DIMMER



For any lamp up to 250 watt. Mounted on switch plate to fit in place of standard switch. Virtually no radio interference. Price 23.35, Industrial model 5A 23.30

10 AMP DIMMER CONTROL

For the control of lighting on stage or in a studio or for control of portable equipment in workshops, etc. This has two 13 amp socket outlets each is controlled by a 5 amp solid state regulator. The overall length is 17 in., width \$\frac{1}{2}\times, and depth \$1\frac{1}{2}\times. In the end is fitted a master On/Off switch indicator, lamp and tuse. Price \$8.25.

-STANDARD WAFER SWITCHES-

Standard size 1½in. water—silver-plated 5 amp. contact, standard ½in. spindle 2in. long—with locking washer and nut.

| No. of Poles | 2 way | 3 way | 4 way | 5 way | 6 way | 8 way | 9 way | 10 wav | 12 way |
|--------------|-------|-------|-------|-------|-------|--------|-------|--------|--------|
| 1 pole | 44p | 44p | 44p | 44p | 44p | 44p | 44p | 44v | 44p |
| 2 poles | 44p | 44p | 44p | 44p | 44p | 44p | 44p | 77p | 77p |
| 3 poles | 44p | 44p | 44p | 44p | 77p | 77p | 77p | £1.04 | £1 04 |
| 4 poles | 44p | 44p | 44p | 77p | 77p | 77p | 77p | £1.32 | £1.32 |
| 5 poles | 44n | 44p | 77p | 77p | £1 04 | £1 04 | £1 04 | £1 60 | £1.60 |
| 6 poles | 44p | 77p | 77p | 77p | £1.04 | \$1.04 | £1.04 | £1.87 | £1.87 |
| 7 poles | 77p | 77p | 77p | £1.04 | £1 32 | £1.32 | £1.32 | £2.15 | £2:15 |
| 8 poles | 770 | 77p | 770 | £1.04 | £1.32 | £1.32 | £1.32 | £2.42 | £2.42 |
| 9 poles | 77 p | 77p | £1 04 | £1.04 | £1.60 | £1.60 | £1.60 | £2.70 | £2.70 |
| 10 poles | 77p | 77p | £1.04 | £1.32 | £1.60 | £1.60 | £1.60 | £3.00 | £3.00 |
| 11 poles | 77p | £1-04 | £1.04 | £1.32 | £1.87 | £1.87 | £1.87 | £3.25 | £3.25 |
| 12 poles | 77p | £1 04 | £1 04 | £1.32 | £1·87 | £1 87 | £1.87 | £3·52 | £3.52 |

DISTRIBUTION PANELS

Just what you need for work bench or lab. 4×13 amp sockets in metal box to take standard 13 amp fused plugs and on/off switch with neon warning light. Supplied complete with 6 feet of heaveable. Wired up ready to work, £2.48 plus 25p P. & P.



CAPACITOR DISCHARGE CAR IGNITION

This system which has proved to be amazingly efficient and reliable was first described in the Wireless World about a year ago We can supply kit of parts for an improved and even more efficient version (Practical Wireless, June). Price \$5.55 plus 30p post. When ordering please state whether for positive or negative systems, Also available, ready made ignition systems for 6v. vehicles. \$5.78 plus 20p.

CENTRIFUGAL BLOWER

CENTRIFUGAL BLOWER
Miniature mains driven blower centrifugal type blower unit
by Woods, powerful but specially built for quiet running—
driven by cushioned induction motor with specially built
low noise bearings. Overall size of blower is approx. 44° x
44° x 4°. When mounted by its flange air is blown into the
equipment but to suck air out mount it from the centre using
a clamp, ideal for cooling electrical equipment, or fitting into
a cooker hood, film drying cabinet or for removing flux smoke
when soldering etc., etc. A real bargain at £2:04.



ZPM MODULATION MOTOR

Could also be used to open ventilators, doors, valve, damper, etc. particularly suitable for remote control. Made by Satchwell. Essentially a reversible geared motor fitted with internal limit switches to stop it at the end of its travel. Size approx. 6in. X 6in. X 5iin. and weighing approx. 10 lb. This is extremely powerful and would lift a heavy door or open a long line of ventilators. To operate this motor you prust the 50 cycle supply through a changeover switch. For instance a thermose, chicken hatchery, etc. An indicator on the motor graduated 0-10 shows the state of open close. Also internally fitted is a variable resistor, wires from this to a volt meter would give a remote indication of the open or close position. A very expensive motor if both direct from Satchwell, our price complete with step down Transformer is £15-15.

MULLARD UNILEX

MULLARD UNILEX
This D.1.Y. Stereo Amplifter is still available complete at \$7 for the four Mullard Modules, or Modules can be bought separately as follows:— 4 watt amplifier module (2 required) Mullard Ref. No. E.P.900, \$1.60.

Pre-amp module Mullard Ref. No. E.P.900, \$2.52 seach.

Power Module—Mullard Ref. No. E.P.900, \$2.52 seach.

In addition and made to Mullard specification we offer:—
Standard Control Unit with securcheon and Knobs. \$3.30. Knobs, Set of 4, 50p. Special offer the complete Unitex with control pane at Pre V.A.T. price, \$10 post paid.

THIS MONTH'S SNIP-

TAPE PLAYBACK UNITS

Mains operated. Made by Reditume the famous "music in background people." These are complete units ready to work and we understand that they are in good going order. We have not tested them but would exchange any that do not work properly. These have a superior motor driven flywheel to control the tape through the capstan and also an even equally useful valve amplifier with ELS4 output. In a steel case with carrying handle. Two models offered, good as new £6.50 and somewhat used at £3.50. 75p carriage up to 200 miles then 50p per 100 miles extra. 100 miles extra





ELECTRIC TIME SWITCH

Made by Smiths these are A.C. mains operated. NOT CLOCKWORK. Ideal for mounting on rack or shelf or can be built into box with 13A socket. 2 completely adjustable time periods per 24 hours, 5 amp changeover contacts will switch dreit on or off during these periods. 22.75 post and ins., 23p. Additional time contacts 55p pair.

MULLARD AUDIO AMPLIFIERS module form, each ready built complete with I sinks and connection tags, data supplied. Model 1183 500mW power output 72p. Model 187 500mW power output 94p. Model RP9000 4 wat power output \$4.60. EP9001 two channel or stereo pre amp. \$1.99. 10% discount if 10 or more ordered.





WINDSCREEN WIPER CONTROL Vary speed of your wiper to suit conditions. Al instructions to make. £2.48. s. All parts and

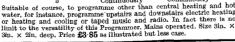
PAPST MOTORS

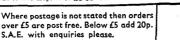
PAPST MOIONS Est. 1/40th h.p. Made for 110-120 volt working, but two of these work ideally together off our standard 240 volt mains. A really beautiful motor, extremely quiet running and reversible. 21.65 each. Postage one 25p, two 33p, 230v. model 23:30.

HORSTMANN 24-HOUR TIME SWITCH

ms this could

| ie | as follows:- | | |
|----|--------------|----------------------|---------------------|
| 1 | rogramme | Hot Water | Central Heating |
| | - 0 | Off | Off |
| | 1 | Twice Daily | Off |
| | 2 | All Day | Off |
| | 3 | Twice Daily | Twice Daily |
| | 4 | All Day | All Day |
| | ŝ | Continuously | Continuously |
| æ | course to | programme other than | central heating and |
| | | | |





NEW ITEMS THIS MONTH

AUTO-TRANSFORMER
Upright mounting, fully shrouded with terminal blocks for input and output. Made by a famous company for imported fridges, 150w. continuous loading or 250w. intermittent. Very nice transformer and probably cost £2 to make. Our Price £1.65 plus 30p post and service.

3 CORE MAINS FLEX
Metric size .5 mm. which is approx. equivalent to the old 14/36 rating. Suitable therefore for mower or similar portable tools. Cores are colour coded to the new European standard. Brown—live . . . Blue—neutral . . Yellow/Green—earth. Grey P.V.C. covered overall. 100m. coils 25, 50m. coils 23 and 25m. coils 21.75. Post 40p on 100m. 25p for 50m. coils and 20p for 25m. coils.

PUSH ON/PUSH OFF PANIC SWITCHES

Tough enough to be foot switches and elegant enough to be panel switches. Bakelite construction with white push rod. Price **40p** less quantity discounts. Switch is rated 250 v.—

20 WATT CAMPING LIGHT

AU WAII CAMPING LIGHT
Also makes good car emergency light. This uses a standard
2 foot 20 watt tube and operates from a 12v. car battery
drawing approx. 1A. This gives illumination per amp, hour
of battery life far in excess to filament lamps and in fact
to the miniature 8-13 watt camping lights often offered.
Complete unit ready to operate, in strong white enamelled
metal frame. These would normally sell at £6, are unused
but slightly solled and we offer these at £3·50 plus 40p
post and packing.

6 MAINS TRANSFORMERS

Our Ref. MTM1. 27 volt at 8 amp. Upright mounting fully shrouded—flying leads—fully tapped primary. Price 23:30. Our Ref. MTM2. 12v. at 1 amp. Upright mounting with fixing lugs, tag connection, 240v. primary—12v. 1 amp. secondary. Price 82p.

secondary. Free 82p.

Our Ref. MTM3, 6.3v. 2 amp, upright mounting with fixing lugs, tag connections, 240v. primary 6.3v. 2 amp, secondary. lugs, tag co Price 77p.

Our Ref. MTM4. 18v. 1 amp. with thermal cut-out, upright mounting with fixing lugs—tag connections. Primary 240v.—secondary 18v. at ½ amp. Price 88p.

Mains Isolation MTM5, 350 watts earth shielded—flex peads—upright mounting lugs for fixing. Price £4.40 each. Maius Transformer Bargain, Standard mains 240v. input Secondary 2.4v. 9 amp. intermittent 5 amp. continuous. Price 55p.

6 VOLT RELAYS

Our Rei. REL M1, Removable clear plastic enclosed two sorew fixing. Have a 6 volt D.C. operated coll but two change-over contacts, each rated to make and break 240v, A.C. at 15 amps. Price 66p. each.

P.O. TYPE 3000 RELAY Our Ref. REL M2. 100 ohms coil and heavy duty contacts. Price 75p each. oil and 2 pair normally open

3 INTERNATIONAL TRANSISTOR

Printed in several languages including English, Carefully compiled and each contains a wealth of information.

PHOTO-MULTIPLIER TUBE

American R.C.A. type No. 4555. These tubes have a gain of a million or more. Regular Price—over £15. We have a limited quantity and offer these at £4.50. Brand New. Not tested, but any not perfect would be exchanged.

6 DIGIT COUNTER
Resettable. 440 ohm coil up to 25 impulses per second.
Ex-equipment but guaranteed perfect. £2.20 each.

TANGENTIAL AIR MOVING UNITS

Work as extractors or blowers especially quiet running, perfectly balanced, plastic impeller—self oiling bearings, induction motor—outlet size 12in. × 2in. approx. Ideal to fit in cooker hood. £2.20 plus 30p post and ins.

DESK TELEPHONES

Ex G.P.O. Black standard model with dialing dial but no internal bell, Supplied with connection diagram 94p each, post etc. 40p for single then 65p per pair. Ditto, with bell but without dialing dial 98p each plus 40p post for single then 65p per pair. then 65p per pair.

Type 14 D.M.G. A very fine switch made by Honeyweil. The switch is intended for mounting on panel through oblong hole. No serves required for fixing, its sprung clips secure it quite firmly. The operating button is approx. Iln, dia. round and dished for ease of operation. Has 2 sets of 10 amp. change-over contacts. Spring loaded, returns to normal when pressure is released. Ideal for instrument or quality gear. Price 50p each. 10 for £4.50. 100 for £40.

American made. Our Ref.: REL Al. Measures only \$\frac{1}{2}\text{in}\$, thick and \$\frac{2}{2}\text{in}\$. Hell Al. Measures only \$\frac{1}{2}\text{in}\$, wide \$x\$ in, thick and \$\frac{2}{2}\text{in}\$. In thick and \$\frac{1}{2}\text{in}\$. In thick and \$\frac{1}{2}\text{in}\$. In thick and \$\frac{1}{2}\text{in}\$. How the contact rating but estimate this at \$\frac{1}{3}\text{5}\text{amps}\$. The coil resistance is 600 ohms and 9-12 voit will close it. Ideal for models and miniaturised equipment, It's a plug in relay but we supply complete with base. Price 28p including base.

COMBINATION SWITCH

COMBINATION SWITCH

This comprises of 12 miniature change-over micro switches. Joined in banks of 3 and mounted on frame with four digital numbered thumb wheels and a removable lever for locking the thumb wheel—the thumb wheel operates 3 banks. Over 4.000 combinations are possible but by "e-wiring the switch connections underneath then thousaids more variations are possible. If you are making equipment which you don't want switched on accidentally or without authority then this is a switch to consider—this can also be used as a coding switch for many other operations. Very neat and compact and measuring approx. 4in. × 14in. × 14in. deep. Priced at £3.03.

MAGNETIC CLUTCH
XEROX 1215494-J/N 10-1110 PN866-10. We have no information on this but it appears that the main section with coil fits to the spindle of the machine and there is a contact plate to fit on a stationary part. It appears also that the clutch can be used as a partial brake by putting reduced voltage into it, as a normal brake with normal voltage or as emergency stop by putting increased voltage into it. American made and very well made at that. Price £1.65.

SUB. MINIATURE MICROSWITCH Made by Burgess, their ref. V476—our ref. MS.Al. These measure only \$in. \times \$in. \times \$in. thick—have change over contacts and tag connection. Price 16p each or 10 for \$1.44.

J. BULL (ELECTRICAL) LTD.

(Dept. W.W.) 7, Park Street, Croydon, CR0 1YD Callers to 102/3, Tamworth Road, Croydon

RVICE TRADING

ALL PRICES INCLUDE V.A.T. POSTAGE AND PACKING. Overseas, please ask for quotation.

MATSUNAGA VARIABLE VOLTAGE TRANSFORMERS INPUT 230 v. A.C. 50/60 OUTPUT VARIABLE 0/260 v. A.C.



.. £107-80 0-260 v. at 50 amps Special discount for quantity I AMP OPEN TYPE (Panel Mounting)
23 | amp £7.70 | 2½ amp £8.86

4 amp £5.23

| L.T. TRANSFORMERS | |
|-------------------------------|-------|
| All primaries 220-240 volts. | , |
| Type No. Sec. Taps | Price |
| 1 30, 32, 34, 36 v. at 5 amps | £5·50 |
| 2 30, 40, 50 v. at 5 amps | £7·92 |
| 3 10, 17, 18 v. at 10 amps | £5·83 |
| 4 6, 12 v. at 20 amps | £7·70 |
| 5 17, 18, 20 v. at 20 amps | £8.58 |
| 6 6, 12, 20 v. at 20 amps | £8·14 |
| 7 24 v. at 10 amps | £6·16 |
| 8 4, 6, 24, 32 v. at 12 amps | £8-47 |
| 9 6 and 12 v. at 10 amps | £4-51 |

36 volt 30 amp. A.C. or D.C Variable L.T. Supply Unit

VALIBIJE L. S. SUPPIY UNIX
Input 220/240 v. AC. Output Concinuously variable 0-36 v. A.C./D.C
Fully isolated. Fitted in robust meta
case with Voltmeter, Ammeter, Panel
Indicator and chrome handles. Input and Output fully fused
ideally suited for Lab. or Industrial use. £77

MOTOROLA MACII/6 PLASTIC TRIAC 400 PIV 10 AMP

Now available EX STOCK supplied complete with full data and applications sheet. Price £1-21. Suitable Diac 22p.

240 V A.C. SOLENOID OPERATED FLUID VALVE

Will handle liquids or gases up to 7 p.s.i. Forged brass body, stainless steel core and spring. \(\frac{1}{2} \) in 5.p. inletfoutlet. Precision made. British info. PRICE: \(\frac{22}{22} \) 22.9. Special quotation for quantity. NEW in original packing.



FOOT SWITCH

Suitable for Motors, Drilis, Sewing Machines, etc., etc. 5 amp. 250 Volt. Price 99p.





New ceramic construction, vitreous enamel embedded winding, heavy duty brush assembly, continuously rated.

25 WATT 10/25/50/100/250/500/1k/1-5k/0.5k/5k ohm

50 WATT 1/5/10/25/50/100/250/500/1k/1-5k/2-5k/5k ohm

£1·54 100 WATT 1/5/10/25/50/100/250/500/1k/1·5k/2·5k/3·5k ohm Black Silver Skirted knob calibrated in Nos. 1-9. 1½ in. dia brass bush. Ideal for above Rheostats, 28p ea.

UNISELECTOR SWITCHES-NEW

4 BANK 25 WAY FULL WIPER 25 ohm coil, 24 v. D.C operation £6.71 6 BANK 25 WAY FULL WIPER 25 ohm

coil, 24 v. D.C. £7-48, 8 BANK 25 WAY FULL WIPER 24 v. D.C. operation £8-69,



Each bank comprises of a change-over rated at 10 amps 240 volt A.C. Black knob 1 in. dia. Fixing hole \$\frac{1}{2}\$ in. Prices: 1-bank 33p, 2-bank 44p, 3-bank 55p. (Illustrated) inc. P. & P. Special quotes for quantities.



VERY SPECIAL OFFER
MICRO SWITCH
5 amp. c/o contacts. Fitted with 5 amp. c/o contacts. Fitted with removable metal plate Ex P.O. 20 for £1 10 (min. order 20).

'HONEYWELL' LEVER OPERATED MICROSWITCH



STROBE! STROBE! STROBE!

FOUR EASY TO BUILD KITS USING XENON WHITE LIGHT FLASH TUBES, SOLID STATE TIMING + TRIGGERING CIRCUITS, PROVISION FOR EXTERNAL TRIGGERING, 230-250V. A.C. OPERATION.

EXPERIMENTERS "ECONOMY" KIT
Adjustable 1 to 30 Flash per sec. All electronic components including Xenon Tube + instructions £7:26

NEW INDUSTRIAL KIT
Ideally suitable for schools, laboratories etc. Roller
in printed circuit. Adjustable 1-80 f.p.s., approx.
i
coutput of Hy-Lyght. Price £12:10.

HY-LIGHT STROBE
Designed for use in large rooms, halls and utilizes a silica tube, printed circuit. Speed adjustable 1-20 f.p.s. Light output greater than many (so called 4 Joule) Light output greater strobes. Price £13-75.

'SUPER' HY-LIGHT KIT Approx. 4 times the light output of our well proven Approx. 4 times the light output of our wen proximally. Lyght strobe. Variable speed from 1-13 flash per sec. Reactor control circuit producing an intense white

ONLY £22-88, incl. P. & P. ATTRACTIVE, ROBUST, FULLY VENTILATED METAL CASE for the Super Hy-Lyght Kit including reflector, £8-25.

FOR HY-LYGHT STROBE incl. reflector, £4.95

7-INCH POLISHED REFLECTOR. Ideally suited for above Strobe Kits. Price 66p.

·,************ RAINBOW STROBE FOUR LIGHT CONTROL MODULE

MODULE
Will operate four of our Hy-Lyght or Super Hy-Lyght
Strobes in either 1, 2, 3, 4 sequence; 2 +; or all together.
Thoroughly tested and reliable. Complete with full
connection instructions. Price: £20-80. Send S.A.E. for
details.

COLOUR WHEEL PROJECTOR

Complete with oil filled colour wheel. 100 watt lamp. 200/240V AC. Features extremely efficient optical system. £20-57, 6 INCH COLOURWHEEL As used for Disco lighting effects, etc. Price £5-72



BIG BLACK LIGHT

400 Watt. Mercury vapour ultra violet lamp. Extremely compact and powerful source of u.v. Innumerable industrial applications also ideal for stage, display, discos etc. P.F. ballast is essential with these bulbs. Price of matched ballast and bulb £18-15. Spare bulb £8-03.

BLACK LIGHT FLUORESCENT U.V. TUBES
4ft. 40 watt. Price £6-38. (For use in standard bi-pin
fluorescent fittings). MINI 9 inch 6 watt black light U.V.
tube. £1-65. Complete ballast unit and holder for 9 in.
tube £2-09.

^*************** **ELECTRONIC ORGAN KIT**



Easy to build, solid state Two full octaves (less sharps and flats). Fitted hardwood case, powered by two penilte 1½v, batteries. Complete set of er with full instructions and

parts including speaker, etc., together 10 tunes. £3·52.

HONEYWELL PROGRAMME TIMERS

240V. A.C. 5 r.p.m. motor. Each cam operating a c/o micro switch. Cams are individually variable, allowing inumerable combinations. Ideally suited for machinery control, automation etc. Also in the field of entertainment, for chaser lights, animated displays, etc.
15 cam model £6-50
10 cam model £5-50



SIMPLE 12 CAM PROGRAMMER with 4 adjustable cams and 8 that may be profiled to individual requirements. Available with 15 or 13 r.p.m. motor £4·18

24 HOUR TIMER

Can be adjusted to give a switching delay of between ½ hr. to 24 hrs. Driven by 200/ 2507. A.C. synchronous motor. 15 amp. c/o contacts. Mfg. Crater Controls Ltd. Supplied with scale calibrated 0-10 (2 hours per division)



VENNER ELECTRIC TIME SWITCH

200/250 volt. Ex-GPO. Tested, perfect condition. Two ON, two OFF, every 24 hrs. at any manually pre-set time. Price: 15 amp, £3-35, 20 amp, £4-39, Also available with Solar Dial ON at dusk, OFF at dawn. Prices as above.





INSULATED TERMINALS Available in black, red, white, yellow, blue and green. New 11p each, incl. P. & P. Minimum order 6.

METER BARGAIN

BALANCE/LEVEL MIETERS

100-0-100 Micro Amp. Size $1\frac{1}{2}$ in. \times $1\frac{1}{2}$ in. \times $\frac{7}{8}$ in. Price only 83p



AMMETERS NEW! 2½in. FLUSH ROUND available as D.C. Amps 1, 5, 15, 20 or A.C. Amps 1, 5, 10, 15, 20. Both types £1 98 incl. P. & P. 0-300V. A.C. £2 98



RELAYS SIEMENS PLESSEY, etc. MINIATURE RELAYS 1 2 3 4 1 2 3 4 4 66p* 700 12-24 2 c/o 66p* 52 4-6 6M 66p* 700 15-35 2c/oHD 77p* 52 4-6 4 c/o 88p* 700 6-12 1 c/oHD 55p* 180 6-12 4 c/o 88p* 700 6-12 1 c/oHD 55p* 180 6-12 4 c/o 88p* 700 6-12 1 c/oHD 55p* 180 6-12 4 c/o 88p* 700 16-24 6M 66p* 200 9-12 2 c/o 77p* 1250 24-36 4 c/o 66p* 410 10-18 4 c/o 77p* 2500 36-45 6 M 68p* 600 9-18 2 c/o 66p* 2400 30-48 4 c/o 66p* 700 16-24 4 M28 66p* 2400 30-48 4 c/o 55p* 700 16-24 4 4 4 6 5 6p* 2400 30-48 5 6 M 68p* 700 16-24 4 4 6 6p* 36p* 300 40-70 2 c/o 55p* 700 16-24 4 4 c/o 88p* 15k 85-110 6 M 55p* (1) Coll ohms; (2) Working d.c. volts; (3) Contacts; (4) Price HD—Heavy Duty. All Post Paid. (*including Base)

Price MD=Heavy Duty. All Post Paid. (rind 12 VOLT D.C. RELAY Type 1: Three sets c/o contacts 5 amp. 88p (Similar to illustration below). Type 2: One set c/o contacts 66p Type 3: 4-8 volt 3 c/o HD, 67 ohm coil. 88p.

SPECIAL OFFER.
700 ohm. 4 c/o Ex. new equipment. £55-90 per 100 incl. bases (minimum 100).

230 VOLT A.C. 'DIAMOND H' RÉLAYS

(Unused)
Three sets c/o contacts rated at 5 amps.
Price 66p. incl. P. & P. (100 lots £44-06)
24 volt A.C. 3 c/o 66p.



230 VOLT A.C. RELAYS
One set c/o contacts rated at 7.5 amps. Boxed. Price 55p.

MINIATURE RELAYS 9-12 volt D.C. operation. 2 c/o 500 M.A. contacts. Size only 1 in. ギメキ in. Price 65p. 30-36 v. D.C. operation. 2 c/o 500 M.A. contacts. 3,200 ohm coll. Size only 1×ホメ非 in. 44p.

MINIATURE LATCHING RELAY
Mig. by Clare-Elliott Ltd. (Type F) 2 c/o permanent latching
in either direction. Coil 1150 ohm. 15-30 v. D.C. New 77p,

INSULATION TESTERS (NEW) Test to I.E. Spec. Rugged metal construction, suitable for bench or field work, constant speed clutch. Size L. 8 in., W. 4 in., H. 6 in., weight 6 lb., 500 VOLTS, 500 megohms £30.80 I,000 VOLTS, I,000 megohms £37.40



230V/240V COMPACT SYNCHRONOUS GEARED MOTORS

Manufactured by either Sangamo, Haydon or Smith. Built-in gearbox.
5 RPM A/cw 3 RPH A/cw 20 RPH cw 2 RPH cw 6 RPH cw 12 RPH cw cw=Clockwise. A/cw=Anti-clockwise



All at 83p

REVERSIBLE SPLIT PHASE MOTOR 250 r.p.m. 100-115/210-240V AC, 2 in. × 1 in. Ideal for rim-drive models, display etc. Extremely powerful for size 83p. (including small capacitor.)

4 BANK 3 c/o PUSH BUTTON ASSEMBLY

Complete with black rectangular bottoms, 5 units £1:10 (5 units





BLOWER UNIT

200-240 Volt A.C. BLOWER UNIT Precision German built. Dynamically balanced, quiet, continuously rated, reversible motor. Consumption 60mA. Size 120mm. dia. x 60mm. deep. Price 23-52.

(Type J) 71 r.p.m. torque 10 lb. in.
Reversible 1/70th h.p. cycle 38
amp. (Type 2) 28 r.p.m. torque 20
lb. in Reversible 1/80th h.p. 50 cycle 28 amp.
The above two precision made U.S.A. moi lb. in Reversible 1,980th h.p. 50 cycle -28 amp.
The above two precision made U.S.A. motors are offered in 'as new' condition. Input voltage of motor 115v A.C. Supplied complete with transformer for 230/24v A.C. input.
Price, either type £4.84 or less transformer £2.75.
These motors are ideal for rotating aerials, drawing curtains, display stands, vending machines, etc, etc.

PARVALUX TYPE SD2. 200/250 VOLT A.C. D.C. HIGH SPEED MOTOR

Speed 9,000 r.p.m. approx. or 3,200 r.p.m. if used with built-in governor, or variable speed over a wide range if used in conjunction with our Dimmer Switch, illustrated below. PRICE: £2:20





600 WATT DIMMER SWITCH

Easily fitted. Fully guaranteed by maker s Will control up to 600 watts of all lights except fluorescent at mains voltage. Complete with simple instructions £3:34

AEL MAIL ORDERS, ALSO CALLERS AT:

57 BRIDGMAN ROAD, CHISWICK, LONDON, W4 5BB. Phone: 01-995 1560 Closed Saturdays.

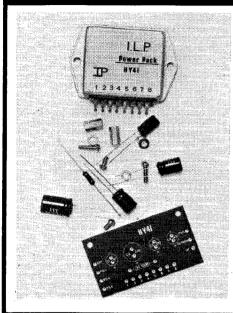
SERVICE TRADING CO. SHOWROOMS NOW OPEN

AMPLE PARKING

PERSONAL CALLERS ONLY

9 LITTLE NEWPORT STREET, LONDON, WC2H 7JJ. Tel.: 01-437 0576





THE HY41

The HY41 supersedes the popular HY40 introduced by ILP last year. This highly improved module achieves true High Fidelity with a dramatic reduction in distortion (typically 0.05% at 1KHz into 8 ohms!) and is electronically and mechanically compatible with the HY40.

With this important improvement the HY41 retains all of the quality characteristics found in the earlier version and P.C. board, Resistor, Capacitors, Hardware Mountings and comprehensive manual are included in the basic kit. No further components are required to construct a complete power amplifier of extremely high performance sufficiently versatile to provide power not merely for Hi-Fi but also for public address systems and industry.

The free manual gives a full circuit diagram of the HY41 and its various applications including a complete stereo amplifier.

Like its predecessor the HY41 is based on conventional and proven circuit techniques developed over recent vears.

OUTPUT POWER: British Rating 40 WATTS PEAK, 20 watts

R.M.S. continuous.

LOAD IMPEDANCE: 4-16 ohms,
INPUT IMPEDANCE: 30K ohms at 1KHz.

VOLTAGE GAIN: 30db at 1KHz

TOTAL HARMONIC DISTORTION: less than 0.15% (typical 0.05%)

FREQUENCY RESPONSE: 5Hz-50KHz + 1db. SUPPLY VOLTAGE: + 22.5volts D.C. SUPPLY CURRENT: 0.8 amps maximum.

PRICE: inc. comprehensive manual, P.C. board, five extra components and P. & P.:-

MONO: £5.39 STEREO: £10.78 This is inclusive of V.A.T. plus P. & P.

UNIQUE HYBRID PRE-AMPLIFIER

The HY5 has rapidly established a position in the WORLD as the sole hybrid pre-amplifier to contain all feedback and equalization networks within an integrated pre-amplifier circuit.

Supplied with the HY5 are two stabilizing capacitors and by the addition of

volume, treble and bass potentiometers it is ready for use.

Internally the HY5 provides equalization for almost every conceivable input, the

Two distinction is achieved by use of a multi-way switch or by direct interconnection,
Two distinctive features of the HY5 are its inbuilt stabilization circuit, allowing it
to be run off any unregulated power supply from 16–25 Volts and a balance circuit
which, when linked by a balance control to a second HY5, forms a complete stereo pre-amplifier.

Specifically and critically designed to meet exacting Hi-Fi standards, the HY5 combines extremely low noise with a high overload capability. When used in conjunction with the HY41 and PSU45 forms a completely intergrated system.

Magnetic Pick-up (within ±1db RIAA curve) 2mV. 47K Ω Tape Replay (external components to suit

tape Replay (external components to suit head). 4mV. $47K\Omega$ Microphone (flat) 10mV. $47K\Omega$ Ceramic Pick-up (equalized and compensatable) 20-2000mV. variable. Tuner (flat) 250mV. $100K\Omega$ Auxiliary 1 250mV. $47K\Omega$ Auxiliary 2 2-20mV. $100K\Omega$

ACTIVE TONE CONTROLS (Bexendall)
Treble ± 12db.
Bass + 12db.

INTERNAL STABILIZATION

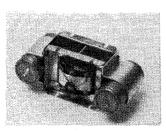
Enables the HY5 to share an unregulated supply with the Power Amplifier. SUPPLY VOLTAGE

16-25 volts

PRICE: MONO: £3.96

SUPPLY CURRENT 6mA approx.
OVERLOAD CAPABILITY better than 26db on most sensitive input infinite on tuner and auxl. **OUTPUT NOISE VOLTAGE: 0.5mV.**

STEREO: £7.92 This is inclusive of V.A.T. plus P. & P.



POWER SUPPLY PSU45

The versatile P.S.U.45 is designed to supply your HY41's +HY5's in stereo or mono format.

Specification

Input: 200–240 Volts.
Output: ± 22.5 Volts at 2 amps.
Overall Dimensions: L. 7"; D. 3.8"; H. 3.1"

PRICE: £4.95 This is inclusive of V.A.T. plus P. & P.

Please note we reserve the right to substitute at our discretion updated versions of advertised designs where applicable

CROSSLAND HOUSE · NACKINGTON · CANTERBURY · KENT

CANTERBURY 63218

DOUBLE BEAM OSCILLOSCOPE

DOUBLE BEAM OSCILLOSCOPE
Designed for investigation and measurement of pulsed and periodic waveforms.
The use of two independent vertical
deflection amplifiers permits the display
and analysis of two different waveforms
simultaneously. Display area 35x90 mm.
Repetition rates of investigated waveforms 50 Hz to 1 MHz. Range of pulse
length 0.35 ps to 1 sec. Range of amplitudes 0.04 to 400V. Maximum amplitude
without external attenuator 100V.

tudes 0.04 to 400V. Maximum amplitude without external attenuator 100V. Characteristics of vertical amplifiers: Amplifier passband at 1db DC to 1 mHz. Amplifier passband at 3db DC to 5 mHz. Sensitivity at medium frequencies at broad passband 500mm/V. Size of undistorted pulse display 40mm. Input Impodance 0.5 + .015 Megohms shunted by 45pF. Input Impedance with strength attenuator 5. Morohome hunted

shunted by 45pF. Input Impedance with external attenuator 5 Megohms shunted by 13pF. Voltage attenuation ratios of the built-in attenuator 1:1: 1:10: 1:100: 1:100. Time Base: Preset calibrated sweep durations:—microseconds per cm 0.2: 0.5: 1: 2: 5: 10: 20: 50: 100. Milliseconds per cm 0.2: 0.5: 1: 2: 5: 10: 20; 50: 100. Pree-running time base frequency range 50 Hz to 1 mHz. Sweep sync. voltage and trigger voltage 0.5V. Maximum trigger pulse repetition rate 10 kHz. Built-in amplitude calibrator: Amplitude of test pulses with duration of 0.35µsec or longer 0.04-100V. Fundamental error of the calibrator

Electronic Brokers



PORTABLE AC/DC RECORDING VOLTAMMETER

Fitted with separate zero-marking pen. Accuracy 1.5% DC. 2.5% AC. Measurements ranges — AC and DC: 5-15-150-250-500MA 1.5-5 Amps 5-15-50-150-250-500V. DC only 150mV. Frequency range 45 to 1000 o/s. Chart width 100mm. Chart speeds 20-60-180-600-1800-5400 mm/hour. Weight 22 lbs. Price complete with accessories. Price complete with accessories

£78.00



4-RANGE **GENERAL** PURPOSE TEM PERATURE BECORDER Type 01

Specially designed compact self-contained instrument for recording temperatures up to 500°C. The main design objectives were for an easy-touse, robust instrument suitable for use, robust instrument suitable for use, robust instrument suitable for one four ranges are 10°C, 50°C, 100°C and 500°C. These are selected by push buttons allowing full use of the 3" wide chart. Two chart speeds 1" and 6" per hour are provided by the 240V 50Hz synchronous chart drive.

The 3% basic accuracy of the instrument, The 3% basic accuracy of the instrument, which is adequate for most applications, has been achieved without introducing stability problems in the d.c. ampliffer, making the recorder ideal for use in schools colleges and universities and by unskilled personnel. The recorder is complete with N1Ch/NiA1 thermocryple and mains lead. This product is brand new and manufactured in our own laboratories with three month guarantee

£95.00

£5.00 packing and carriage.



OUTPUT METER TYPE MU 964

This instrument basically consists of a transistorized amplifier voltmeter which measures the voltage across a specified load. It is provided with 40 load values ranging from 2.50hm to 20K0hm As the loads are purely resistive, their value keeps constant with varying frequency. A special negative feedback loop allows a nearly linear scale to be obtained. No damages to the instrument result from errors in presetting the load values or the power ranges.

Power measuring range (in 4 ranges) Level mea: Ref. 1mW Frequency range Accuracy Load input resistances

Pacietances accuracy

Instrument Calibration

from 1mW to 10 W from -3 dB to +4 from 20 Hz to 50KHz

Within 0.5 dB 40 Value: better than 5%

£89



WHEATSTONE BRIDGE AND CABLE FAULT LOCATOR

Measurement of resistance in the range of 0.005 to 1 megohm. Location of cable faults using Varley loop method. Location of cable faults using Murray loop method. Measurement of asymmetry of wires. Use of four-decade section as a resistance. Use of four-decade section as a resistance box. The bridge consists of four decade switches giving a range from 1 to 9999 ohms in 1-ohm steps. Accuracy: from 1 to 99,990 ohms 0.5%, from 0.1 to -.9999 ohms 1.5%, from 100k to 1 to -.9999 ohms 5.0%, from 0.005 to 0.0999 ohms 5.0%. Dimensions: 300x230x 150 mm. Weight: Approx. 12 lbs. Price complete with connecting leads

£41.00

THREE CHANNEL

HIGH SPEED RECORDER

£90.00

with accessories



AC/DC MULTIMETER

With taut band suspension movement. Sensitivity 20,000 ohms per volt on DC and 4,000 ohms per volt on AC.

and 4,000 ohms per volt on AC. Technical Data: 0.06-0.6-6-60-600mA-3 Amps DC. 0.3-3-30-300mA-3 Amps AC. 0.6-1.2-3-12-30-60-120-600 DC. 1200 Volts 3-6-15-60-150-1300-600-900 Volts AC. 45 to 20,000 Hz. 5000, 5-50-500k @ resistance. Decibel range —10 to +12dB. Accuracy (% of F.S.D.):—DC and resistance measurements ±2.5. Price with test leads, and storage case

£8.00



3" SINGLE BEAM PULSE OSCILLOSCOPE

For display of pulsed and periodic wave ror uispay of pulsed and periodic wave-forms in electronic circuits. Vertical amplifier: Bandwidth 10 MHz. Sensi-tivity at 100 KHz v RMs/mm.1-25. Horizontal Amplifier: Bandwidth 50 KHz. Sensitivity at 100 KHz v RMs/mm. Free running 20-200,000 Hz in .3-25. Preset triggered sweep $1-3000\,\mu$ nine ranges. Calibrator pips. Dimensions $220x360x430\,$ mm. Weight 40 lbs. 115-230V AC operation.

£39.00



10 CHANNEL **EVENT RECORDER**

Designed for recording sequences of up to ten different operations, e.g. sequence of machine tool operation, switching sequences, etc. Record is presented in the form of square "pulses". When energised, pen moves by approximately 4mm. to the right of zero line. Response time 100 milliseconds. Chart width 110mm. Chart length 50ft. Inv. capacity 72 hours. Chart speeds 20-60-180-600-1800-5400 mm/hour. Size 160x160x255mm. Weight 9 lbs. Price complete with accessories Strip Chart Recorder. Chart length 175ft. Footage indicator. Width of recording channel 80mm. Chart speeds (selected by pushbuttons) 1.2-12-30-60-120-300-600-3000 mm. per minute. Full deflection current 8mA. Internal impedance 210 ohms. External impedance 800 ohms. Dimensions 510x345x175 mm. Weight 44 lbs. Price complete with accessories

£52.00

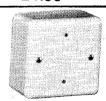
£10.50



MILITIMETER

0.1-1-10-100-1000mA 2.5-10-20-250-500-1000V AC/DC Sensitivity AC and DC all ranges except 10V-10,000 Ohm/V. Dimensions 212x118x75 mm, Weight 2.9 lbs. Price complete with steel carrying case and test leads

£4.95



ELECTRONIC TIME DELAY SWITCH.

Specification:

Delay period 1-25 minutes adjustable, load 1000 watts maximum. Operating Voltage 180-250V a.c. 50Hz. Siza 32 x 32 x 12 standard lovy Surface mounting Box. Trade Price £5.80

NEW AMPERTEST 690—

Electronic Brokers Limited announce the introduction of a new clamp type ammeter having 6 current ranges plus 2 voltage ranges for use on 50 to 60 alternating current supplies. Known as the AMPERTEST 690, the new instrument is manufactured in Italy by Industria Construzioni Ellectromeccaniche, one of Europe's largest manufacturers of electrical measuring instruments.

The Ampertest 690 uses the familiar clamp or 'pincer' system to measure the current flowing in a conductor without breaking the circuit. The meter, which is designed to be used with one hand, has 6 current ranges from 3 to 600 amps f.s.d. with the first division at 100 mA. The current ranges may be extended by use of a 10 to 1 current transformer which is supplied with the instrument providing ranges from 300 mA to 60 amps f.s.d. with the first division at 10 mA. In addition there are two a.c. voltage ranges, 250V and 600V f.s.d. The connections for voltage measurements are made by two leads and probes which plug into the base of the instrument.

instrument.

The range to be used is selected by rotating a small serrated thumbwheel on the side of the instrument. This action brings the appropriate scale under the meter needle removing the possibility of reading the wrong scale. When not in use the meter movement is clamped by the ON-OFF switch to prevent

matter movement is callified by the damage during transit.

The Ampertest 690 is supplied complete with voltage measuring leads and probes and combined twin wire adaptor/current transformer in a solid leather carrying case. A belt fitting pouch is available as an extra.

£37.50





Measurement ranges:—Current 10-25-100-250-500 Amps. Voltage 300, 600 V. Accuracy 4%. Scale length 60mm. Overall dimensions 283x94x36mm. Weight 1.5 lbs.

c value in Test Equipm

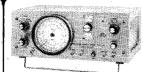


MODEL 300 LOGIC PROBE

A compact easy-to-operate logic probe. As a light-emitting diode is used the unit actuates with low power. It does not affect the circuit under test because of high input impedance. Up to as high a frequency as 12 MHz

£5.50

plus 75p. packing and carriage.



TV SWEEP MARKER **GENERATOR Type VU 167**

Suitable for alignment of tuned circuits in television sets. Incorporates a sweep generator. a market generator and a crystal-controlled oscillator operating at 5.5 MHz. Sweep frequency range 1-30 MHz, 170-260 MHz Fund, 470-780 MHz Harmonic. Marker frequency range 2-266 MHz, 480-800 MHz



PORTABLE WHEATSTONE BRIDGE

Designed for measurements of DC resistances in the range of 10-5 to 106 Megohms. Basic accuracy 0.01 ohms to 10° ohms is 0.2%. Dimensions 300x230x 150 mm. Weight approx. 13 lbs. Supertester 680R is a completely new concept in measuring instruments. In itself a high quality test meter with eighty ranges on a 128mm minror backed of the appropriate accessories it can be.

Accessories to convert the supertester 680R to a complete measurement system. With the addition, and the propriating guess measure a wide range of values of the appropriate guess and phase sequence. And there system offers many advantages over conversatility and economy.

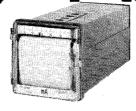
Signal Temperature

Temperature

Temperature

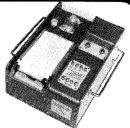
Temperature

The Supertester 680R is a completely new concept in measuring instruments. The passive set of the appropriate accessories it can be addition. The positional results and phase sequence. And there system offers many advantages over conventional test meters including versatility and economy.



MINIATURE PEN RECORDER

Provides permanent record of DC currents up to 1mA. Eminently suitable for use where space is limited. Separate time marker pen provided. Chart width 80mm. Chart length 40ft. Chart speeds: Slow 20-60-180 mm/hour. Fast 600-1800-5400 mm/hour. Dimensions 120x 120x285mm. Weight 7.7 lbs. (3.5 Kg).



SINGLE CHÂNNEL HIGH SPEED RECORDER

Chart length 175ft. Footage indicator Width recording channel Chart speeds (selected by push buttons) 1.2-6-12-30-60-120-300-600-3000 mm per minute. Full deflection current 8mA. Internal impedance 210 ohms. External impedance 800 ohms. Dimensions 320x340x175mm. Weight 35 lbs. Price complete with accessories

£55.00

Amperclamp //

For measuring from 250r 500 amps £11.95

Gauss Meter For measuring nagnetic ield

strengths

signals for circuit testing £5.95 Phase Sequence

Indicator To indicate the phase a 3 pha

supply. £5.95.



range — 30 to + 200°C. £11.95

Electronic Voltmeter



input resistance of 11Mohms for d.c. and 1.6Mohms shunted by

Transistor

£11.00.

Tester

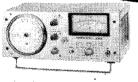
For tran-sistors and

£18.50 Complete with case & probes

SUPER TESTER 680R Specification

13 D.C. ranges from 0.1 to 2000V. 12 to ranges 50,4A to 20,000 from 5A. Q/A 1% Accuracy 1%
11 A.C. ranges
2 to 2500V. 10 ranges from

Resistance: 6 ranges from 0.52/to 100M2 Reactance: 1 range of 0-10M. Frequency: 2 ranges of 0-50 OTHER ACCESSORIES AVAILABLES HUNTS
D.C., 25, 50 and 100 amps. £4.50 each.
CURRENT TRANSFORMERS A.C. 25 and 100 amps. £7.00 each. E.H.T. PROBE Extends
CURGENT TRANSFORMERS A.C. 25 and d.c. voltage to 25,000v £5.95. Accuracy



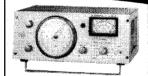
DISTORTION METER Type D 566 B

Fully transistorised for measurement of rully transistorised for measurement of overall distortion of signals with frequencies between 10 Hz and 1 MHz. Built-in electronic voltmeter can also be used separately for measuring AC voltage, basic noise, gain or attenuation over a wide frequency range. Distortion meter:

Distortion meter:—
Frequency range (in 5 ranges): from 10
Hz to 1 MHz. Distortion factor (in 7 ranges): from 0.03% to 100. Minimum testing voltage: 300 mV approx. Input impedance: 100 KOhm; 40 pF approx. Millivoltmeter:

Voltage range (in 12 ranges); from 1 mV Voltage range (in 12 ranges): from 1 m to 300 V f.s.d. Level range (rel. to 0.776 V): from + 52 dB to - 75 dB. Frequency range: from 10 Hz to 2 MHz. Bandwidth (within 3 dB): up to 8 MHz. Accuracy: better than 5%. Input impedance: 2 MOhm; 50 pF approx.

£249.00



LF SIGNAL GENERATOR Type G 1165 B

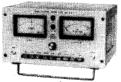
Transistorised generator providing wide range of squarewave and sinewave signals. Suitable for measuring distortion, gain or attenuation when testing the frequency response of low-frequency equipment.

Sinusoidal output:-

Sinusoidal output:— Frequency range (in 4 ranges): from 10 Hz to 100 KHz. Output voltage: from 1 mV to 10 V. Output impedance: 600 Ohm constant. Frequency accuracy: better than 2%. Harmonic distortion: less than 0.3% (50 Hz. . . 30 KHz).

Frequency range (in 4 ranges); from 10 Hz to 100 KHz. Output voltage; from 100 mV to 10 Vp. Output impedance: to 10 Vp. Output impedance: Ohm constant. Risetime: less than

£165.00

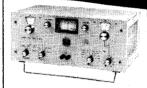


WOW AND FLUTTER METER Type WF 971

Solid state, high stability unit. Can be preset for either the European standard at 3150 Hz or the American standard at Hz. Provided with built-in oscillator.

oscillator. DIN and CCIR. Input Signal: 20 mV rms to 20 V rms approx. Frequencies (switchable): 3150 Hz and 3000 Hz. Ranges (flutter): +/- 0.1% +/- 0.3% +/- 1% f.s.d. Drift indication: +/- 2% max. Input impedance 10m0hm max. Built-in oscillator 10m0hm max. Built-in oscillator 3000 Hz or 3150 Hz switchable. Stability: better than 0.1%, Shifts for calibration: +/- 0.1% dynamic, 50 Hz + 2% static.

£225.00



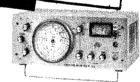
RCL BRIDGE Type P 966

For measurement of RCL and capacitor dissipation factor and inductors figure of merit Q. Consists of a system of switchable bridges. a 1 KHz generator, and a sensitive tuned detector. Particularly suitable for testing of small production batches, and selection of companions. batches and selection of component parameters.

. Measurement ranges

Measurement ranges: Resistance: from 0.1 Ohm to 11 MOhm Capacitance: from 1 pF to 1100 μ F Inductance: from 10 μ H to 1100 H Accuracy: +/-1%. Dissipation factor D from 1.10-3 to 50. Quality factor Q from 1.10-3 to 50. Quality factor Q from 0.02 to 1000. Internal oscillator

£170.00



AM-FM GENERATOR Type AF 1065

Permits fast and accurate calibration of modern radio receivers. Suitable for calibration and testing in the laboratory. AM frequency range: from 140 KHz to 46 MHz in 6 ranges expanded range 9.5-12 MHz; 85-110 MHz. Frequency range: 9.5-12 MHz; 85-110 MHz. Frequency occuracy: better than 1%. RF output voltage: adjustable from 0.1 µV to 0.1V. Output impedance: 75 Ohm constant. Modulation: AM; FM; AM + FM. Amplitude modulation: 400 Hz; from 0.50% adjust. Frequency modulation: 1000 Hz adjust. Deviation from 0 - +/- 50 KHz. External modulation: AM; FM; from 30 Hz to 15 KHz. Permits fast and accurate calibration of

£225.00



WIDE RANGE TRANSISTOR AUDIO GENERATOR

High stability low frequency generator. Basic circuit is a Wien-Bridge controlled sine wave oscillator and square wave is produced by means of a Schmitt trigger circuit. FREQUENCY RANGES: 4 from 10Hz to 100Hz. DUTFUT VOLTAGE: 1 millivolt pp to 1 volt pp ± 3% for sine and square wave. IMPEDANCE: 1K ohm. DISTORTION FACTOR: 0.2% for the sine wave under the power frequency range. 2 10% for output for the lower frequency range, < 1.0% for output for the lower frequency range. < 1.0% for the upper frequency range. RISE TIME: < 0.1 microseconds for the squere wave and 0.3 microseconds in the upper frequency range. WORKING VOLTAGE: 9 volt, 240mm x 140mmH x 100mmD. 2\frac{1}{2}\text{lbs.}£27.



WIDE RANGE TRANSISTORISED R.F. SIGNAL GENERATOR RF carrier can be internall and externally modulate

and externally modulated from 0.5% — internal modulation frequency is variable between approximately 800 and 1200 ftz — AF output adjustable for level and frequency FREQUENCY RANGES: 8 from 147kHz to 220.0 MHz. SCAEE ACCURACY: 2% or ± 0.1% when set against internal crystal calibrator. RF OUTPUT VOLTAGE: 50 millilvolts minimum on all ranges. AM MOGULATION: Internal and external from 0 to 50%. AF OUTPUT: Variable for frequency and level. WORKING VOLTAGE: 9 Volt. 240mml x 140mmHx 100mml0.3†lbs.£36.



WIDE RANGE TRANSISTORISED R.F. SIGNAL GENERATOR

A few priced R.F. Generator, 8 ranges, covers from A low priced R.F. Generator, 8 ranges, covers from 150kHz in 300MHz. The output can be unmodulated by means of an internal 400Hz oscillator. The A.F. signal is available from a separate output socket Both output sockets are provided with an isolating capacitor (500 D.C. max) FREUENCY. 42% R.F. OUTPUT VOLTAGE: 50 millivolts minimum. A.F. OUTPUT VOLTAGE: Approximately 1 Volt at 800Hz. WORKING VOLTAGE: 9 Volt. 240mml x 140mml x 100mmd0. 2‡lbs. £19.50



TRANSISTORISED CAPACITY/RESISTANCE

The measuring ranges of this Capacity Resistance Bridge are 10pF to 100mfd capacity and 10ohms to 10 megohms. Bridge balance is shown by means of a luminescent belance indicator tube. A further witch position permits measurement of capacity leakage current. RESISTANCE RANGES: 3 from 10 ohms to 10 megohms. CAPACITY RANGES: 3 from 10pF to 10mfd. WORKING VOLTAGE: 9 volt. 240mmL x 140mmH x 100mmD. 2 lbs. £18.50

OSCILLOSCOPE -- TYPE 46

OSCILLOSCOPE—TYPE 46
A compact general purpose single beam, solid state, DC
Y Amplifier — 5mHz. 3 db bandwidth D.C. coupled dc — 5mHz. 100mV — 50V.
9 steps, Maximum input (a.c. coupled)
300Vp. Input impedance approx
1M2& 47pf. Sweep speeds ±5%. 4 ranges 10mSec/cm to 10Sec/cm

Trigger/Sync

Automatic operation.

3db bandwidth D.C. impedance approx. 2.5 k \$\(\) 220-240v 50-60Hz. Width 24cm. Depth 19cm.

ALL EQUIPMENT BRAND NEW AND GUARANTEED FOR 6 MONTHS

ADD 10% VAT TO ALL PRICES ● PROMPT DESPATCH MAIL ORDER CALLERS WELCOME MON-FRI 9 A.M. to 5-30 P.M. SAT 9-30 A.M. to 2 P.M.

towards the cost nack carriage on all items and ing U.K. (except where aiready packing and carriage . are indicated);

LECTRONIC BROKERS LTD. 49-53 Pancras Road, London NW1 20B. Telephone 01-837 7781

PEN RECORDERS

Portable 1, 2 and 4 channel pen recorders by Kelvin Hughes. General purpose recorders providing clear instantaneous and permanent records of phenomena with comparatively high rates of change. The torsion-strip suspension of the moving coil renders the instrument immune to the effects of vibration and accoleration.

tion.

Six possible chart speeds, chart width 55mm., length 150 ft., linearity 8 v. at 3 mA. response D.C. to 100 c/s.

Single pen with amplifier £49 4 pen with amplifier £49





SERVORITER Model FWS

SERVORITER Model FWS

By well-known American manufacturer.
Power supply 120 v. 50 Hz. Response time
24 secs. Resistance source 10 K ohms max.
Chart width 11 in. This is a slow-speed recorder that can be used for measuring any
quantity with a comparatively slow rate of
change such as temperature, humidity etc.
Supplied with electrovoit controller that
enables the sensitivity, reset, proportional
band and rate to be adjusted. This unit enables
the demanded temperature to be controlled
and the actual temperature recorded. Size:
10½ in. wide, 17½ in. high, 13½ in. deep. Price
£95. Carriage extra.



MUTUAL INDUCTANCE BOX TYPE R.7005

Specification Range: 0-11.100 mH. in 0.002 mH. divisions. Accuracy: ±(0.3 × 0.012M/ where M=value of mutual inductance in mH. set on the box. Frequency range: 0-2.5 Kc/s for all decades except X1=0-15 Kc/s. Maximum current: 0.5A for decades 1 A for variometer (both primary and secondary wingings). Case: Polished teak. List price £65. Our price £25-30.



SINE COSINE POTENTIOMETER 47K

Precision component by Pye, Model 2002.
Manufactured to rigid Ministry specification. The assembly consists of three units mounted in one frame. Each unit contains two sine and two coside potentiometer sections, the sliders being ganged together. Electrical connections, 2 end taps, slider and centre tap, Mechanical I/P: 30 r.p.m. Max. torque: 3½ o.2./in. Dimensions: W. 6½ in. H. 5 in. D. 7½ in. Wt. 7½ lbs. Ex equipment. Good condition. Price £5. Carriage extra.



ROHDE & SCHWARZ Zg DIAGRAPHS HF IMPEDANCE PLOTTER

These instruments will rapidly plot the loci of the impedance or admittance of any item such as antennas, transformers, absorbers, filters and other networks. Impedance measurements are possible from 0-02Z, to 50Z, where 20=50, 60 or 75 ohm.



75 ohm. Type ZDD 300—2400MHz. Price **£390.**

INFRA-RED SPECTROPHOTO-METER

A single beam instrument designed primarily to analyse the effluent from a gas chromatograph, however the fast response and fast scan capabilities make it suitable for fast reaction studies involving conventional gas, liquid, or solid samples. The wave length range is 2·5 to 14·5 microns. In excellent condition.



EX COMPUTER POWER SUPPLIES Input 220/250 Volts Output 22 Volt 10 Amp Overload Protect Lamp. £6-50

EX COMPUTER HIGH GRADE FULLY STABILISED POWER SUPPLIES

Input 200/250V. ADVANCE TYPE DC 207

ADVANCE TYPE DC 200

ADVANCE TYPE DC 202

ADVANCE TYPE DC 197

WESTINGHOUSE Fully Fused Input 200/220/240/10

WESTINGHOUSE

Output 20 Volts 9 Amps, 10 Volts 5 Amps.

-20 Volts 13 Amps. 10 Volts 5 Amps. 20 Volts 2·5 Amps. 35 Volts 9 Amps. 24 Volts 4 Amps. 10 Volts 8 Amps.

6 Volts 7-5 Amps -6 Volts 11 Amps. -28 Volts 9 Amps.

14 Volts 0.75 Amps. 20 Volts 4 Amps. 25 Volts 2.5 Amps. 30 Volts 0.75 Amps.

£18 EACH. P. & P. £2.

KENT CHROMALOG 1 DIGITAL INTEGRATOR

For use with gas chromatography apparatus or anything with an output expressed as a varying direct voltage. Automatic print out and 0-10mA O/P to drive recorder. Offered in excellent condition, 3 months warranty and copy of handbook. Price £150. Carriage extra.





K.G.M. Type 3015F 7 Segment display showing figures 0-9 plus decimal point. Character of 9mm height. In 16 DIL case.

NEW LOW PRICE £1-85

SN7447N BCD Decoder Driver £1.05

COMPUTER PARTS

TELETYPE PUNCH

TELETYPE PUNCH
BRPE High-speed punch. Selfcontained, consists of punch
unit, base, motor unit. For use in
many data communication systems. Operating speeds up to
100 characters per second. (1100
words per minute). Available for
punching 5, 6, 7, or 8, level codes,
into \(\frac{\psi}{2}\)" and 1" tape. Synchronous, parallel-wire input.
£195.

WELMEC 7 & 8 HOLE FLEC-

TRO-MECH-ANICAL PUN-CHES & READER Models S110 and R82C, 17 char. per sec. Rebuilt, avail-able from stock. £75.

Magnetic Tape Transporters AMPEX TM4, TM2, TM7, FR300, IBM 7330, POTTER, ICL Magnetic Drums. From £75.

HAND PUNCHES-80 COLUMN



THE PUNCH

Is a table-mounted Model Punch. For the Serial Punching or alpha-numeric Data, Alpha or Multi-Hole Punching. From £59-50. 103 Verifiers available £89-50.

ICT KEYBOARDS in original packing—Numerical from £4.50

ICT KEYBOARDS

original packing — Alpha-neric **Prices From** from £12·50

IBM PUNCH CARD EQUIPMENT FULLY GUARANTEED

Α

| 026 056 | Automatic alphanumeric keypunch | 820 00 380 00 |
|------------|---------------------------------|------------------|
| | Carriage extra. | |

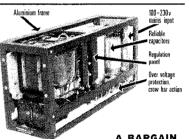
MARCONI A.M. SIGNAL GENERATOR TYPE TF801D

10-485Mc/s in five ranges. Output 0-1μV-1 Volt E.M.F. External Sine A.D. Frequency 30c/s-50Kc/s. PRICE £195



SOLARTRON Type OS103 L.F. DECADE OSCILLATOR

Frequency range: 0-01Hz to 11-1KHz. Sine and square wave accuracy ±2% O/P 0-10V r.m.s. 100mW into 1KΩ PRICE: £99-00



A BARGAIN
IN NEW POWER SUPPLIES.
IS THAN HALF MANUFACTURERS PRICES. AT LESS

O/P Voltage 7.5V-9V. Max. load current 10 Amps. Max. ripple on full load approx. 60mV. p.p. Threshold current 10.5A. Overvolt protection. OUR PRICE £12.50

SIGNAL GENERATOR "X" BAND

Sanders model CT480 (SG480) and CT478 (SG478). Specifications: CT480 8-11-8KMHz; CT478 1-3-4-2 KMHz. OIP of ImW from 8-0-11-0KMHZ (CT480) and 1-4-4-0 KMHZ (CT478). These high grade generators comprise a klystron oscillator in a co-exial cavity fed from a stable power source. Provision for application of square wave or pulse modulation, internal or external sources. Attenuator calibrated from 0-100 db below 1mW. I/P 110-250V 50-500Hz 200W. Rack mountling. W. 19in. H. 14In. D. 15in. Wt. 74 lbs. Price. £125.

AUTOMATIC RECORDING WAVE ANALYSER

Muirhead Type K100. The K100 Automatic Recording Wave Analyser can analyse and record graphicalit complex waveforms in the range 10c/js to 19kC/s. The speed of analysis makes it possible to reduce testing times to a minimum, resulting in increased utilisation of personnel and reduced running costs of equipment. Specification: Frequency Range 10c/s-19kC/s in four ranges. Frequency Accuracy ±1% on recorder paper (1-5% bandwidth, steady frequency signal services of the recorder paper (1-5% bandwidth, steady frequency signal).

10c/s-19Kc/s in four ranges. ±1% on recorder paper (1.5% bandwidth, steady frequency sig-

Frequency Accuracy

andwidth, steady frequency signal).

Selectivity

1.5% bandwidth, 3dB down to approx. 0.8% off-tune. Octave discrimination better than 70 db, *5% bandwidth 3 db down to approx. 2.5% off-tune. Flat-topped with 1 db over 3% frequency range. Octave bandwidth 3 db down to approx. 12% off-tune. Octave discrimination 3 db down to approx. 12% off-tune. Octave discrimination 30 db*.

* Final scope of response curve is 6 db/Octave above and below tune frequency.
Further information available on request.

PRICE £450

AIRMEX 853 H.F. WAVE ANALYSER

Frequency range: 30KHz-30MHz in 7 bands. A very popular instrument which has been refurbished. Price £45.

| ٧ | OMET | ERS | | |
|----|---------|-------------|-----------------|-----------|
| | Model | Price | Model | Price |
| | D | £14 | 40 | £16 |
| | 7 | £19·50 | 47.A | £16 |
| | 8 | £29·50 | 48A | £16 |
| | 8X | £35 | | |
| в. | 47A and | 48A are Adm | iralty versions | s of Mode |

NB. 47A and 48A are Admiralty versions of Model 40, the only difference being that the resistance ranges 2-1·2, 0-12 ohms, which are available on the Model 40 with the use of an external power supply, are not available on the 47A and 48A. CASES AND LEADS EXTRA

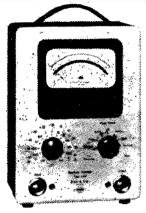
PHILIPS VALVE VOLTMETER MODEL GM6014

Max. 300mV, 1000Hz-30MHz.

PRICE £30.00



to purchase some of the World's finest calibration instruments a



BRUEL & KJAER **ELECTRONIC** VOLTMETER **MODEL 2489**

An instrument for the measurement of peak average peak average absolute and true R.M.S. values of AC voltages in AC voltages in the frequency range 2 Hz - 200 KHz. 11 voltage ranges ranges from mV - 1000 vo 10 mV - 1000 volts, Full scale deflec-tion. Built-in ref, voltage. Input im-pedance 10 Meg. The high input volte

impedance, low output impedance and accurate attenuator also nake this an ideal calibrated amplifier. Brand new in manufa turer's original cartons.

List Price £98-50 OUR PRICE, BRAND NEW COMPLETE WITH ACCESSORES £59-50

7-TRACK DIGITAL MAGNETIC TAPE STORAGE DECK

These machines, originally ex-computer, are multi-track recording units, ideal for data storage. Record and Replay Heads encased in one common unit. Low resistance heads. Frequency response approximately 0 Kc/s to 50 Kc/s. Bit density 557 b.p.l. ½ in., 102 in. spools, 230 V to 380 V Capstan motor speed 1,500 r.p. 48 V DC rewind motors complete with vacuum assembly. Finished in brush aluminium and matt black. Size 27 in. x 26 in. x 8 in. Weight 90 lbs. Price £72-50.

MEMORY PLANES



Ferrite core memory planes with wired Ferrite cores. Used for building wired refrite cores. Used for building your own computer or as an interesting exhibit in the demonstration of a computer. Mounted on plastic material frame 5 x 8 in. Consisting of matrices 40 x 25 x 4 cores, each one indiducely addressable, and divided into 2 halves with independent sense

and inhibit wires. Price £2.50.

EAC DIGIVISOR Mk. II DIGITAL READ-OUT DISPLAY

Ideally suitable for use in conjunction with transistorised decade counting devices. No need for amplifiers or relays as only a few milliwatts of power are required to charge the digits. The DIGI-VISOR incorporates a moving coil movement which moves a translucent scale through an optical system and the resultant single plane image is projected on a screen. The translucent scale is made to represent digits 0-9. Specification: 6·3 Volt, 250 Microamp. Image height 7/8 in. Size 4 1/16 x 2 7/64 x 1 5/8 ir. List price £8-90. Our Price £2.

LOW FREQUENCY RESOLVED COMPONENT INDICATOR BY SOLARTRON

Type VP 253.2A. This instrument will indicate by means of two centre zero 6 in. scale meters the resolved components of a signal voltage with respect to the applied reference energisation. Frequency range 0.5 c/s-1 Kc/s. Signal voltage ranges: 50 MV, 150 MV, 500 MV, 1-5V, 5V, 50V, 15V, and 150V, with either balanced or unbalanced input. Signal input Resistance: 10 M2 unbalanced, 20 MΩ balanced. Reference input Voltage 90/130 or 230/240V. Standard Rack Panel 19 in. x 12½ in. New condition complete with manual. Price £45-00.

VHF ADMITTANCE BRIDGE

Wayne Kerr B801A 1-100 MHz. Conductance 0-100 milliohms. Capacitance 0-230 pF and 0-230 pF £25-00. Also B901. Indicates parallel components of conductance and positive or negative capacitance for lines, antennas and feeders 0-100 milliohms, 0 to +/- 75 pF and -75 pF. Accuracy 2% up to 250 MHz. Price £75-00.

FENLOW LO' / FREQUENCY ANALYSER

0.3 Hz to 1 KHz. Power density 0-10. Bandwidth switching range 06: 0.3: 1.5: 7.5: 37.5 Hz. Price £95.

HONEYWELL ELECTRONIC POTENTIOMETRIC RECORDER

The following types are available:

1. Model Y15301115-01-01-0-(150)-01-002-202. Range 600 SYNCHROVI 1300°C.R. PRICE £95-00.

2. Model Y153X18-/VAH/-11-111-118-/P8/DN2 Range 0-20 MV. PRICE £95-00 PRICE £95-00

BRAND NEW MINIATURISED STRIP CHART RECORDER BY DUSTRAK

A CAUSTRAK

of America. This Recorder indicates the magnitude of applied currents or voltages by a continuous distortion-free line on pressure sensitive paper. Moving coil movement, scale calibrated 0-1 milliamp d.c. internal resistanct.

100 ohms. Chart drive motor 240 V 59 Hz. Chart speed 1" per hour. Complete with handbook. Price £35.00 plus £5.00 packing and carriage.



SINGLE PEN RECORDER

RECORDER
by Record Electrical. 3" chart, sensitivity 1 milliamp, chart speed 1" and 6" per hour. Size 8" x 11" x 6". Offered complete with per assembly and spare chart. Listed at over £100 — this month's special price due to bulk purchase . . . £39.50 plus £5.00 packing and carriage.

TRANSITROL TEMPERATURE CONTROLLER TYPE 990

CONTROLLER TYPE 990

Completely transistorised self-contained direct deflecting units for indicating and controlling temperature accurately over a wide range. Sultable where a signal can be converted into d.c. Sensitivity 10 ohms per My. Minimum F.S.D. 8 MV. Cold junction compensation. Calibrated scale length 8.5°.

-880°C. Accuracy +1—4%. Front panel size 10° × 8½°, weight 11 lbs. Mains supply 100-250 V. Control switching and thermocouple connections all at back of case. Price £18.50 plus £2.00 packing and carriage.

POWER SUPPLIES, IBM EX-COMPUTER HIGHLY STABILISED, TRANSISTORISED LOW VOLTAGE POWER SUPPLIES.

These modular units'incorporate overload protection on bott INPUT and OUTPUT. Load regulation of 1% or better. Lov rippie and fest response time. Input voltage 120-130 50 Hz available in the following types:

6 volt 8 Amp £21.00 6 Volt 12 Amp £17.00 12 Volt 16 Amp £20.00 12 Volt 4 Amp £22.00 12 Volt 12 Amp £22.00 12 Volt 20 Amp £24.00 30 Volt 7 Amp £19.00





VIBRON ELECTRO-

WIBRON ELECTROMETER MODEL 33B

An exceptionally stable laboratory instrument for the measurement of very small d.c. voltages and currents derived from a high impedance. The Vibron Electrometer has input ranges of 10 MV. 30 MV. 100 MV. 30 MV. and 1 V and the loon all ranges. The drift does not 42 hours and the input resistance.

output is 1 mA. Full scale on all ranges. The drift does no exceed 100 microvolts in 12 hours and the input resistance is 10 to the power of 13 ohms. Price £45.00 plus £5.00 packing

SPECIAL OFFER

ELLIOTT SINGLE PEN RECORDER



A most versatile pen recorde producing a trace on a curvi-linear 3 in. strip chart. Two synchronous speeds: 1 in. and 6 in. per hour.

synchronous speeds: 1 In. and 6 in. per hother. Fitted with high and low alarm contacts operated by the moving coil. Basic movemed only a coil resistance 400 ohms. Fitted with rectifier to allow operation on AC effective coil impedance at 50Hz 1800 ohms. Pgwer supply required.

230V 50Hz. Applications: Ideal for recording relatively slow changing phenomena such as:
Temperature: Gas or liquid flow Rates, Sound Levels, Speed variations. Power Demand, Rainfall, humidity, etc.

Clockwork version also avail-

ASCOP DIGITAL ENCODERS Type 504A-8-001 Price £20. Type EDD8@Price £20.
SYNCHROVERTER SWITCH TYPE G1280

FREQUENCY CONVERTER MODEL B.40

50 KVA to 60 Hz power frequency converter. Fully overhauled. Specification:

Electric Motor 220/380V 50Hz 3ph

Input: 220/3604 30/12 3ph Output: 220V 60Hz 3ph at 50 KVA with PF of 0.8. PRICE £450.00

HEWLETT PACKARD DIGITAL RECORDER MODEL 565A

Data Entry, parallel to 11 columns, Print speed 5 lines per second.

HEWLETT PACKARD SAMPLING OSCILLO-SCOPE MODEL 185B

Including 187C. DC-100 Meg Hz

PRICE £395-00

PYE HIGH RESISTANCE OHMMETER MODEL 10B

Range from 0-3-20,000 Megohms in 4 ranges at 500V. Used for the measurement of components or circuits having high parallel capacitance.

PRICE £20:00



MULLARD VALVE VOLTMETER MODEL E7555/2

PHILIPS SQUARE WAVE GENERATOR MODEL GM2314

tange 15 c/s-200 Kc/s. Duration of square wave pulses between 0.75 /sec and 40 m/sec. Square wave voltage 10V PRICE £75-00

A.E.I. POTENTIOMETRIC RECORDER

0-2-5 MV, 5, 7-5, 10, 15 MV. Chart speed 1 rpm-3 rpm OR 1 rph-3 rph. PRICE £45-00

FOSTER CHART RECORDER TYPE 3671 RY-6 Sensitivity 0-20 MV, -50 MV, 100 MV. Chart speed 1.5 cm/hr-6 cm/hr. PRICE £35-00

DUPLEX RECORD 2 PEN RECORDER day clockwork -50/0/50 m/A

DUPLEX RECORD 3 CHANNEL PEN RECORDER 0-10 M/A. Speed 1" per hour. PRICE £95 00

COLVERN DIGITAL CODERS (Shaft Digitisers) Digital Coders are electromechnical devices, which give a unique parallel digital code output representing the angular position of the shaft. The current handling capacity is sufficient to operate relay decodes and indicators direct without intermediate stage of amplication.3 size magslip, 256 divisions, max. torque for reflected binary code 4-5 oz. ins.

PRICE £10-00.

LEEDS & NORTHRUP STRIP CHART RECORDER
This well-known instrument is fitted with a Series 60 control unit servo
amplifier 101041 BR EQ. Range: 5-571 to 18-855. Ref. junction 320F.
Primary element; P1. P1. 12% RH JMC. Response time: 5 secs. for 1.s. 0.
Chart width: 7 in. Chart speed: 1 in. per hour. Power supply: 120V 50 Hz
(auto-transformer available). Dimensions: Ht. 18", width 11", depth
12½". Weight 51 lbs.

PRICE £120-00

WIDE RANGE OSCILLATOR TYPE 400C by DAWE

FANS BY PLANNAIR

ase 400 c/s-11,000 rpm. Type 1PL41-234

DOUBLE PULSE GENERATOR TYPE TF 1400/S
10 c/s-100 Kc/s. Complete with TM 6600. Pulse adjustable between
1-5 μsec. before and up to 3,000 μsec. PRICE £145-00

R.C. OSCILLATOR TYPE G432 by FURZEHILL Square and sinewave. 250 Kc/s. PRICE £25 0

MARCONI TF867 STANDARD SIGNAL GENERATOR



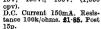
GNAL
Carrier Frequency.
Range: 15Kc/s30Mc/s in 11 bands.
Calibration Accuracy: ± 19.
Stability: After warm
up the driff in a 10minute period is,
typically, less than
0-005% for carrier
frequencies up to
3-2Mc/s and less than
0-01% from 3-2-230Mc/s.
Output Voltage:
0-4µV-4V.
Impedance: 75 ohms
nominal for outputs
from 2-4 v. 75 ohms
for outputs from
4µV-2V. 13 ohms
for outputs from
0-4µV-4V. 3-10Mc/s ±0-5dB or

Accuracy: below 3Mc/s $\pm 0.25dB$ ot $\pm 0.1\mu V$. 3-10Mc/s $\pm 0.5dB$ or $0.2\mu V$. 10-30Mc/s $\pm 1.0dB$ or $\pm 0.5\mu V$. Dimensions: 18 in. high \times 21 in. wide \times 14½ in. deep. Price £165:00



TS60 POCKET MULTIMETER

High-precision at low-cost. Ranges: D.C. 15V., 150V., 1,000V. (10,000 opv). A.C. 15V. 150V., 100V. (1,000





AUDIOTRONIC MODEL
Top value 1000 p.p.v. pocket
multimeter.
Ranges: 0/10/50/250/1000v. AC
and DC. DC Current 0-1mA/
100mA. Resistance 0/1508
On DC Surrent 0-1mA/
100mA. Surrent



LT601 MULTIMETER



New style 20,000 o.p.v. pocket multimeter, 5 / 25 / 50 / 250 / 2509V. D.C. 10/50/100/500/1000V. A.C. 50µA/250mA. 6K/6 meg ohms. —20 to +22db. 23.75, Post 20p.

MODEL TH-12

20,000 o.p.v. Overload protection. Slide switch selector. 0/.25/2.5/10/50/250/1000V.

D.C. 0/10/50/250/1000V. A.C. 0/50µA/25/250mA D.C. 0/3K/30K/300K/3 meg. -20 to +50db. 24.97. Post 15p.



RUSSIAN 22 RAN'
Model U437 10,000 o.p.v.
A first class versatile
instrument manufactured
in U.S.S.R. to the highest
standards. Ranges: 2-5/
10/50/250/500/1000v
D.C. 2-5/10/50/250/500/
1000v A.C. D.C. Current
100 wA/11/0/100 mA/1A.
Resistance 300 ohms/
3/30/300K/3MΩ Complete with batteries, test
leads, instructions and leads, instructions and stordy steel carrying



Our Price **£5**·**97**. Post 25p.

HIOKI MODEL 730X

30,000 O.P.V. Overload protection. 6/30/60/300/600/1200
VDC. 12/60/120/600/1200
VAC. 60 \(\alpha \) A/30 mA/300 mA.
2K/200K/2 megohm. —10 to +63 db. **26**:50. Post 15p.



MODEL PL436

20kΩ/Volt D.C. 8kΩ/Volt A.C. Mirror scale, ·6/3/12/ 30/120/600V D.C. 3/30/ 120/600V A.C. 50/800µA/ 60/600 mA. 10/100K/ 1 Meg/L0 meg Ω. -20 to, +46db, £6·97, Post 12p



MODEL 500

30,000 O.P.V. with overload protection, mirror scale, 5, 5, 2, 5, 10, 25, 10, 10, 25, 10, 10, 25, 10, 10, 25, 10, 10, 25, 10, 25, 10, 25, 15, 25, 15, 25, 15, 25, 15, 25, 15, 25, 15, 25, 15, 10



"JOSTY KIT" HIGH QUALITY

| 1 (| CONSTRUCTION KIT | S |
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| 1 | 1ppointed stockists at all branch | e8 |
| AF20 | Mono Transistor Amplifier | £4.80 |
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| | HF310 or HF325 | £9.96 |
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| | Bands I, II and III | £1.77 |
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| I | AF310 | £21 27 |
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| NT10 | Power Supply 100 m/a 9V Stal | , |
| | 12V Unstah | 26.15 |

Postland Packing 15p per kit
Complete with comprehensive, easy to follow
instructions and covered by full guarantee.

U4312 MULTIMETER



HINKI MODEL 700X

H10KI MODEL 700X
100,000 0.PV. Overload protection. Mirror scale. :3:-61.-21.-518/6/
12/30/601/20/300/600/1200 VDC.
1-5/3/6/12/30/601/150/300/600/1200
VAC. 15/30VA/3/6/30/800/150/300
mA. 6/12 AMP DC. 2K/200 K/2
Meg/20 megohm. —20 to +63db.
\$13.50. Post 20p.



MODEL C-7080 EN

MODEL C-7080 EN Giant 6in. mirror scale. 20,000 c.p.v. |
20,000 c.p.v. |
0/25/12/5/10/50/250/1000/5000V. D.C. |
0/25/25/10/50/250/1000/5000V. A.C. |
0/504A/1/10/100/500mA/10 amp. D.C. |
0/2K/200K/20 meg. —20 to +50 db. #13 95. Post 35p.



370 WTR MULTIMETER

JOW IR MULTIM
Features A.C. current ranges.
20,000 o.p.v.
0/15/2-5/10/50/250/500/1000 V DC
0/2-5/10/50/250/500/1000 V AC
0/50/LA/1/10/100MA/1/10 Amp
DC.

DC. 0/100MA/1/10 Amp AC. 0/5K/50K/500K/5MEG/50MEG. 20 + 62 db. 215. Post 25p.



KAMODEN 72,200 MULTITESTER

High sensitivity tester. 200,000 o.p.v. Overload pre-tection, Mirror scale. Ranges 0/-06/-3/3/30/120/600/1200\.

0/-06/-3/3/30/120/600/1200\
D.C.
0/3/12/60/300/11,206V A.C.
0/6μΑ/1-2mA/120mA/600m/
12A. D.C.
0/12A. A.C.
-20 to 16/37

-20 to +63dB. 0/2K/200K/2 meg/200 meg £16.95. Post 30p.



TMK LAB TESTER

TMK LAB TESTER
100,000 O.P.V.: \$\frac{1}{2}\$ in. Scale
Buzzer Short Circuit Check.
Sensitivity: 100,000 O.PV
D.0. 5K/Voit A.C. D.C.
Voits: 5, 2.5, 10, 50, 250,
1,000V. A.C. Voits: 3, 10,
50, 250, 500, 1,000V. D.C.
Current: 10, 100\(\text{L}\) A. 10,
100, 500mA. 2.5, 10 amp. Resistance: 1K, 10K,
100K, 10MEG, 100MEG.
Decibels: -10 to +49 db. Plastic Case with
carrying handle. Size 7\frac{1}{2} \times 3\frac{1}{2}\$. \$\frac{2}{2}\$. \$\frac{1}{2}\$.


MODEL S-100TR MULTIMETER/ TRANSISTOR TESTER

100,000 o.p.v. MIRROR SCALE/OVERLOAD PRO-TECTION.

0/·12—·6/3/12/30/120/600 V DC.

V DC. 0/6/12/601/22/100 V DC. 0/6/36/12/600 V AC. 0/6/36/12/600 V AC. Ac. 0/12/600 V AC. Amp. DC. 0/10K/1 MEG/100 MEG. —20 to + 50 db. 0-01 —-2 mfd. Transistor tester measures Alpha, beta and Ico. Complete with batteries, instructions and leads. 213-50. Post 25p.



Tests PNP or NPN transistors. Audie indication. Operates on two 1.5v batteries. Complete with all instructions etc. £4.50. Post 20p.



LB3 TRANSISTOR TESTER

Tests ICO and B. PNP/ NPN. Operates from 9v. battery, Complete with all instructions etc. 23.95. instruction Post 20p.



KAMODEN HM.350 TRANSISTOR TESTER

High quality instrument to test Reverse Leak current and DC current Amplification factor of NPN, PNP, ransistors, diodes, SCR's etc. 4in. × 4½in. clear scale meter. Operates from internel batteries, Complete with instructions leads and carrying handle. £12.50 Post 30p.



MODEL 449A IN CIRCUIT TRANSISTOR TESTER

Checks true A.C. beta in/out. Checks Icbo. Checks diodes in/ out. Checks SCR etc. Beta HI 10-500. LO2-50.

LO2-50. Icbo 0-5000µA. 220/240V. A.C. operation, £17-50, Post 25p.



MODEL U4311 SUB-STANDARD MULTI-RANGE VOLT AMMETER

Sensitivity 330 ohms/ Volt A.C. and D.C. Accuracy 5% D.C. 1% A.C. Scale length 165mm.

165mm. 0/300/750µA/1·5/3/ 7·5/15/30/75/150/300/ 750mA/1·5/3/7·5 Amp. D.C. 0/3/7·5/15/30/75/ 150/300/750mA/1·5/3/ 7·5 Amp. A.C.

TE-65 VALVE VOLTMETER

High quality instrument with 28 ranges.

D.C. volts 1.5-1,500 v.

A.C. volts 1.5-1,500 v.

Resistance up to 1,000 megohms a.v. votts 1.5.1,500 v. Resistance up to 1,000 megohms. 220/940v. A.C. operation Complete with probe and instructions. 217-50. Post 30p. Additiona, Probes available: R.F. £2.12 H.V. £2.50.

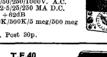


KAMODEN HM.720B F.E.T. V.O.M.

Input impedance 10 meg. ohms. Ranges: 0/25/1/2-5/10/50/250/1000V, D.C.

0/2·5/10/50/250/1000V. A.C. 0/25μΑ/2·5/25/250 MA D.C. -20 to +62dB 0/5K/50K/500K/5 meg/500 meg

£14.95. Post 30p.



T.E.40 HIGH SENSITIVITY A.C. VOLTMETER

10 meg. input 10 ranges: o1/o3/1/s/1/s/1/s/10/s0/100/300 v. R.M.S. 4 cps.-1.2 Mc/s. Decibels -40 to +50 dB. Supplied brand new complete with leads and instructions. Operation 230 v. A.C. £17-50 Carr. 25p.



TMK MODEL II7

FLECTRONIC

VOLTMETER

Battery operated, 11 meg input. 26 ranges. Large 4½ n. miror scale. Size 5½ ×4½ ×2½ n. DC Volts 0.3-200V. AC Volts 3-300V RMS. 8-0-800V. P.P. DC Current -12-22M Ahm. Decibels —20 to +51 db. Complete with leads/instructions. \$17-50. P. & P. 20p.



MODEL L-55 FET V.O.M. Input impedance 10 meg.

ohms. 0/-3/1-2/6/30/120/600V. D.C. 0/-3/1-2/6/30/120/600V. A.C. 0/3/12/60/120/600V. A.C. 0/120μA/120mA. D.C. 0/1K/100K/10 meg/100 meg ohms. £15-97. Post 25p.



KAMODEN HMG-500 INSULATION RESISTANCE TESTER

Range 0-1000 Megohms, 500 Volt.
Battery operated.
Wide range clear
meter 4½in. × 4in.
Complete with de
luxe carrying case,
batteries, instructions, £19-95.
Post 30p.



BELCO AF-5A SOLID STATE SINE SQUARE WAVE C.R. OSCILLATOR Sine 18-200,000 Hz; Square 18-50,000 Hz

Output max. + 10 dB (10 K ohms). Opera-tion internal batteries. Attractive 2-tone case 7½in. × 5in. × 2in. a. × 5in. × Price **£17.50** Carr. 17p.



CI-5 PULSE OSCILLOSCOPE

SCILLOSCOPE
For display of pulsed and
periodic waveforms in electronic circuits. VERT's.
AMP. Bandwidth 10MHz.
Sensitivity at 100KHz.
VRMS/mm. 1-25; HOR.
AMP. Bandwidth 500KHz.
VRMS/mm. 3-26; Presettriggered sweep 1-



triggered sweep 13,000µsec.; free running 20-200,000Hz in nine ranges. Calibrator pips. 220 × 360 × 430mm.
115-230V. A.C. operation. 239.00. Carr. paid.

TO-3 PORTABLE OSCILLOSCOPE 3in. TUBE

3 in. TUBE

Y amp. Sensitivity. Iv
p-p/CM. Bandwidth 1.5 eps
—1.5 MHZ. Input imp.
2 meg Ω. 25 PF. X amp
sensitivity. 9v p-p/CM.
bandwidth 1.5 eps—800
KHZ. Input imp. 2 meg Ω
20 PF. Time base. 5 ranges
10 cps—300 KHZ. Synchronization. Internal/external. Illuminated scale.
140 × 215 × 330 mm. Weight 15 jibs. 220/240 V.
A.C. Supplied brand new with handbook
247-50, Carr. 50p.



RUSSIAN CI-16 DOUBLE BEAM OSCILLOSCOPE

5 mc/s Pass Band. Separate YI and Y2 amplifiers, Rec. tangular 5tn, x 4in. C.R.T. Calibrated triggered sweep from 2 µ/sec, to 100 milli-sec, per cm. Free running time base 50 c/s-1 mc/s. Bullt-in time base calibrator and amplitude calibrator. Supplied complete with all accessories and instruction manual. 287 Carr. paid.



MODEL AT201 DECADE ATTENUATOR

ATTENUATOR
Frequency range:
0-200KHz.
Attenuator: 0-111db.,
0-1db. step.
Impedance 600 ohms.
Max. input power
30dbm.
Size 180 × 90 × 55mm. £12.50. Post 37p.

ARF-300 AF/RF SIGNAL GENERATOR

All transistorised compact, fully portable.
AF sine wave 18Hz. to 220KHz.

AF square wave 18Hz. to 100KHz,

Output sine/square 10v. P.P. RF 100KHz. to 200 MHz. Output 1v. maximum. Operation 220/240v. A.C. Complete with instructions and leads. £29.95. Post 50p.



TE-20 D RF SIGNAL
GENERATOR
Accurate wide range signal
generator covering 120 Kc/s500 Mc/s on 6 bands. Directly
calibrated Variable R.F.
attenuator, audio output.
Xtal socket for calibration.
220/240V. A.C. Brand new
with instructions. £15. Carr.
370. Size 140 × 215 × 170
mm.





230v/240v SMITHS SYNCHRONOUS GEARED MOTORS

Built-in gearbox. All brand new and boxed. 30 RPH CW; 2 RPH CW; 20 RPH CW; 2 RPH ACW; 30 RRH CW. 50p each. Post 12p.



230 VOLT A.C. 50 c/s RELAYS

Brand new, 3 sets of changeover contacts at 5 amp rating. 50p each. Post 10p (100 lots 240). Quantities



ALL PRICES ARE EXCLUSIVE OF 10% V.A.T.

£2 25

"YAMABISHI" VARIABLE VOLTAGE TRANSFORMERS

Excellent quality at low cost, All models—Input 230v. 50/60 c/s. Variable output 0-260v.



MODEL S-260 GENERAL PURPOSE BENCH

| ı Amp | £7.00 |
|---------|--------|
| 2.5 Amp | £8.05 |
| 5 Amp | £11.75 |
| 8 Amp | £15.90 |
| 10 Amp | £22.50 |
| 12 Amp | £23.60 |
| 20 Amp | £49.00 |
| 25 Amp | £58.00 |
| 40 Amp | £82.50 |
| | |



1111 iiii Packing Extra

AUTO TRANSFORMERS

|)/115/230V. Step | up or step | down. Fully shrou | de |
|------------------|------------|-------------------|----|
| 80 W | £2·10 | P. & P. 18p | |
| 150 W | £2.70 | P. & P. 18p | |
| 300 W | £3.60 | P. & P. 23p | |
| 500 W | £5·25 | P. & P. 33p | |
| 1000 W | £7.50 | P. & P. 38p | |
| 1500 W | £10·20 | P. & P. 43p | |
| 2250 W | £17·25 | P. & P. 50p | |
| 5000 W | £35·00 | P. & P. £1 | |

MCA. 220 AUTO-MATIC VOLTAGE STABILISER

Input 88-125 VAC or 176-250VAC. Output 120VAC or 240VAC. 200VA rating. \$11.97. Carr. 50p.



PS.200 REGULATED P.S.U.



Solid state. Variable output 5-20 volt D.C. up to 2 amp. Independent meters to monitor voltage and current. Output 220/240 V. A.C. Size 7½ * × 5½ * × 3½*. \$19-95. Post 25p.

PS.1000B REGULATED P.S.U.



Solid state. Output 6-9 or 12 V. D.C. up to 3 amps. Meter to monitor current. In-put 220/240 V. A.C. Size 4" × 34" × 64". £11.97. Post 25p.



240° WIDE ANGLE IMA METERS MW1·6 60mm. square \$3.97 MW1·8 80mm. square \$4.97 Post extra.



POWER RHEOSTATS

POWER RHEOSTATS
High quality ceramic construction. Windings embedded in vitrous enamel. Heavy duty brush wiper. Continuous rating-Wide range available ex-stock. Single hole fixing, Jim. dis. 5 WATT. 10/25/50/100/250/500/1000/1500/2500 or 000 ohms. \$21.35. P. & P. 10p. 1000 wATT. 1/5/10/25/50/100/250/500/1000 or 500 ohms. \$21.95. P. & P. 15p.

HOMER INTERCOMS



CLEAR PLASTIC PANEL METERS

USED EXTENSIVELY BY INDUSTRY, GOVT. DEPTS., EDUCATIONAL AUTHORITIES, etc.

Over 200 ranges in stock—other ranges to order. Quantity discounts available. Send for fully illustrated brochure.

TYPE SW.100 100 × 80mm



| 100-0 500µ 1mA 20V. 50V. 300V 1 am 5 am 300V | A D-100μA A D.C D.C D.C D.C p. D.C p. D.C p. D.C | £3 95 £3 70 £3 60 £3 60 £3 60 £3 70 £3 70 |
|--|--|---|
| | . A.C deter | £3.70 £4.30 |

| TVDE | SD. | 440 | 62.5mm | v | 85 | mm | From | t. |
|------|-----|-----|--------|---|----|----|------|----|

TYPE SD.460 46mm × 59.5mm Fronts

| 50μA | £2 80 | 500mA | £2.60 |
|-------------|--------|-----------|-------|
| 50-0-50μA | £2·80 | 1 amp | £2.60 |
| 100µA | £2.75 | 5 amp | £2.60 |
| 100-0-100цА | £2.75 | 10 amp | £2 60 |
| 200μΑ | | 5V. D.C | £2.60 |
| 500μA | £2.55 | 10V. D.C | £2.60 |
| | £2.60 | 20V. D.C | £2.60 |
| 1mA 5mA | | 50V. D.C | £2.60 |
| | | 300V. D.C | £2.60 |
| 10mA | £2.60 | 15V. A.C | £2.70 |
| 50mA | | 300V. A.C | £2.70 |
| 100mA | \$2.60 | VU Meter | £2 90 |

"SEW" EDGWISE METERS TYPE P.E.70

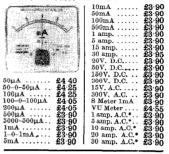


| 3 17/32in | . × 1 15/3 | 2 in. $	imes$ 2 $\frac{3}{4}$ in. deep. | |
|--|------------|---|----|
| 50µА 50-0-50µА 100µА 100-0-100µА 200µА | £3.60 | 500μA 1mA 300V. A.C VU Meter | £3 |

* MOVING IRON-ALL OTHERS MOVING COIL

Please add postage

TYPE MR.85P 44in. × 44in. fronts.



TYPE MR.52P 28in, square fronts.

| 50μA | £3 50 | 10V. D.C £2.50 |
|-------------|-------|---------------------|
| 50-0-50μA | £3.05 | 20V. D.C £2·50 |
| 100μΑ | £3.00 | 50V. D.C £2.50 |
| 100-0-100µA | £2·95 | 300V. D.C £2.50 |
| 500μA | £2·65 | 15V. A.C £2·60 |
| 1mA | £2·50 | 300V. A.C £2.60 |
| 5mA | £2.50 | S Meter 1mA . £2.60 |
| 10mA | £2.50 | VU Meter £3.60 |
| 50mA | £2.50 | 1 amp. A.C.* £2.50 |
| 100mA | £2.50 | 5 amp. A.C.* £2.50 |
| 500mA | £2.50 | 10 amp. A.C.* £2.50 |
| 1 amp | £2 50 | 20 amp. A.C.* £2.50 |
| 5 amp | £2.50 | 30 amp. A.C.* £2.50 |

TYPE MP 4FP Stin v Stin fronts

| I I PE MIN.OSP SEM. A SEM. HOMES | | | |
|--|--|--|--|
| 10V. D.C | £2.60 £2.60 £2.60 £2.60 £2.80 £2.80 £2.80 | | |
| 50mA A.C.*. 100mA A.C.* 200mA A.C.* 500mA A.C.* 1 amp. A.C.* 5 amp. A.C.* 10 amp. A.C.* 20 amp. A.C.* | £3.70 £2.60 £2.60 £2.60 £2.60 £2.60 £2.60 £2.60 | | |
| | 10V D.C. 20V D.C. 50V. D.C. 150V. D.C. 150V. D.C. 15V. A.C. 150V. A.C. 150V. A.C. 300V. A.C. 300V. A.C. 500V. A.C. 500V. A.C. 100MA A.C.* 100MA A.C.* 100MA A.C.* 500MA A.C.* 5 amp. A.C.* | | |

"SEW" EDUCATIONAL METERS



TYPE ED.107

Size overall 100mm \times 90mm \times 108mm.

A new range of high quality moving coil instruments ideal for school experiments and other bench applications. 3in, mirror scale, The meter movement is easily accessible to demonstrate internal working.

| Available in th | e following | ranges: | |
|-----------------|-------------|--------------|------|
| 50μA | | 10V D.C | £9.9 |
| 100µA | | 20V D.C | £5.9 |
| 1mA | | 50V D.C | £5.9 |
| 50-0-50µA | | 300V D.C | £5.9 |
| 1-0-1mA | | Dual range | |
| 1A D.C | | 500mA/5AD.C. | £7 0 |
| 5A D.C | | 5V/50V. D.C. | £7.0 |
| | | • | |
| | | | |

TYPE MR. 38P 1 21/32in, square fronts.

150m A

200mA



200μA 500μA 500-0-500μA

1mA 1-0-lmA ...

10mA 20mA 50mA 100mA

| Sportday | 2 amp | £2.25 |
|--------------|-------------|-------|
| 3.00 | 5 amp | £2.25 |
| 28 | 10 amp | £2 25 |
| 188 : | | |
| · | 3V. D.C | £2 25 |
| Marian III | 10V. D.C | £2.25 |
| . £2·55 | 15V. D.C | £2.25 |
| . £2.50 | | 20 00 |
| | 20V. D.C | £2.25 |
| £2·45 | 50V. D.C | £2.25 |
| £2-40 | 100V. D.C. | £2.25 |
| £2.25 | | 20 00 |
| * \$5.55 | 150V. D.C | £2.25 |
| . £2·25 | 300V. D.C | £2.25 |
| £2·25 | 500V. D.C | £2·25 |
| | 750V. D.C | £2.25 |
| . £2·25 | | |
| . £2·25 | 15V. A.C | £2:30 |
| . £2·25 | 50V. D.C | £2.30 |
| . £2.25 | 150V. A.C | £2.30 |
| | | |
| £2.25 | 300V. A.C | £2.30 |
| £2.25 | 500V. A.C | £2.30 |
| £2.25 | S Meter 1mA | £2.30 |
| | | |
| £2 25 | I VU Meter | £2.65 |
| | | |

| TYPE | MR.45P | 2in, square fronts | ١. |
|--|--|--|---|
| 50μA 50-0-50μA 100μA 100-0-100μA 200μA 500μA 500μA 5mA 10mA 50mA 100mA | £2.70 £2.65 £2.50 £2.50 £2.40 £2.40 £2.40 £2.40 | 2in. square fronts 5 amp. 10V. D.C. 20V. D.C. 30VV. D.C. 50V. D.C. 15VV. D.C. 15VV. D.C. S Meter 1mA VU Meter 1 amp. A.C.* 5 amp. A.C.* | £2.40 £2.40 £2.40 £2.40 £2.40 £2.50 £2.50 £2.50 £2.40 |
| 500mA 1 amp | £2·40 | 20 amp. A.C.* 30 amp. A.C.* | £2:40 £2:40 |
| | | | |

"SEW" BAKELITE PANEL METERS

TYPE MR.65 31in, square fronts.

1 amp. £2.60 5 amp. £2.60 15 amp. 30 amp. 50 amp.

£2.60 £2.60



| 2 C C C C C C C C C C C C C C C C C C C | Of marrie 1111 mile of |
|---|------------------------|
| | 5V. D.C £2.60 |
| **** | |
| | |
| | 20V. D.C £2·60 |
| BEN MANAGEMENT | 50V. D.C £2 60 |
| | |
| | 150V. D.C £2.60 |
| | 300V. D.C £2.00 |
| *************************************** | |
| 25μA £4·60 | 30V. A.C.* . £2.65 |
| | 50V. A.C.* . £2:65 |
| 50μA £3 55 | 150V. A.C.* £2.65 |
| 50-0-50μA £3·05 | |
| 100μA £3·00 | 300V. A.C.* . £2.65 |
| | 500mA A.C.* £2.60 |
| 100-0-100µA £3·00 | |
| 500μA £2.70 | 1 amp. A.C.* £2 60 |
| | 5 amp. A.C.* £2.60 |
| | 10 amp. A.C.* £2.60 |
| 1mA £2.60 | |
| 1-0-1m A £2·60 | 20 amp. A.C.* £2 60 |
| | 30 amp. A.C.* £2.60 |
| | |
| 10mA £2.60 | |
| 50mA £2.60 | VU Meter £3.65 |
| 9011LA 20 00 | 50 mV D.C £2.90 |
| 100mA £2 60 | |
| 500mA £2.60 | 100mV D.C £2.90 |
| | |

5mA . 10mA 50mA 100mA

| ITPES | |
|----------------------------|--------------|
| 50µА 50-0-50µА 100µА | £3.4 £3.4 |
| | 1 |

| . az 00 | 1 TOOM . D.O | |
|----------|-----------------|-------|
| S.80 80m | m Square Fronts | 3 |
| £3·50 | 100-0-100µA | £3.30 |
| . £3·40 | 500μA | £3.05 |
| £3.40 | 1mA | |
| mmmg | 20V. D.C | £3.00 |
| 1 | 50V. D.C | £3.00 |
| | 300V. D.C | £3.00 |
| | 1 amp. D.C | |
| | 5 amp. D.C | |
| | 300∇. A.C | |
| | VU Meter | |
| | | |

UNR 30 RECEIVER



covering 550 kc/s-30 mc/s. nilt-in Speaker 220/240 v. A.C.

OUR Carr. £15.75 PRICE 37p

UR-1A RECEIVER

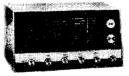


4 Bands covering 550 kc/s-30 FET, S Meter. Variable BFO for Built-in Speaker, Bandspread, tivity Control. 220/240 v. A. 12 v. D.C. 12in. × 4in. × Brand new with instructions.

Carr.

NUR £25.00 PRICE

LAFAYETTE HA-600 RECEIVER



General coverage 150-400 kc/s. 550 kc/s-30 mc/s. FET front end, 2 mech. filters, product detector, variable B.F.O., noise limiter, S Meter Bandspread. RF Gain. 15tn. × 9\frac{1}{2}tn. × 9\frac{1}{2}tn. 18 lb. 220/240 v. A.C. or 12 v. D.C. Brand new with instructions.

£50.00 PRICE

Carr. 50p





Industrial quality in robust metal cases. Battery operation. Volume and squelch controls. Call button and press to talk button. Telescopic aerial. Complete with carrying cases.

2 channel £52.50 Pair. Post 50p.

3 channel **£79.50** Pair. Post 50p. Note: Licence required in U.K.

RECEIVER



SKYWOOD CX203

Send SAE for list of

Semi Conductors and Valves

Solid state, 5 bands covering 200-420 KHz and .55 to 30 MHz, Illuminated slide rule dial. Bandspread. Aerial runing. BFO, AVC, ANL, "S" meter. AM/OW/SS B, Integrated speaker and phone socket, 220/240 v. A.C. or 12 v. D.C. Size 325 × 266 × 150 mm. Complete with instructions and circuit

OUR £32.50 PRICE

TRIO 9R59DS RECEIVER



4 bands covering 550 kc/s to 30 mc/s continuous and electrical bandspread on 10, 15, 20, 40 and 80 metres. 8 valve plus 7 diode circuit. 4/8 ohm output and phone jack. SSB-CW. ANL. Variable BFO. 8 meter. Sep. bandspread diāl. 1F frequency 445 kc/s. audio output 1.5w. Variable RF and AF gain controls 115/250 v. A.C. Size: 7in. x 13in. x 10in. with instruction manual.

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> **ALL PRICES ARE EXCLUSIVE OF 10% VAT**

(RADIO)

EMI LOUDSPEAKERS



Model 350, 13in. × 8in with single tweeter/cross-over. 20-20.000 Hz with single tweeter/cross-over. 20-20,000 Hz. 15 watt RMS. Available 8 or 15 ohms. £7-25 each. P. & P. 37p. Model 450, 13ln. × 8in. with twin tweeters/crossover. 55-13,000 Hz. 8 watts RMS. Available 8 or 15 ohms. £3-62 each. P. & P. 25p.

SPECIAL OFFER! STEREO



SPEAKERS Matched pair of bookshelf speabooksheir spea-kers. De luxe teak veneered finish. Size 14½in. × 9in. × 7½in. 8 ohms. 8 watts RMS. 16 watts peak. Complete with DIN lead.

UILB PRICE £12-95 Carr.

HA-10 STEREO HEADPHONE AMPLIFIER



stereo headphone outputs and separate volume controls for each channel. Operates from 9 v. battery. Inputs 5MU/100MU. Output 50MW.

PRICE

£5·97

MP7 MIXER **PREAMPLIFIER**



NIIR PRICE

P. & P. £8-97

1021 STEREO LISTENING STATION



gain controls, spe switch, stereo he 6in. × 4in. × 21in. OUR

£2.25 PRICE

FM TUNER CHASSIS



OR HIGH QUALITY 6 TRANSISTOR HIGH QUALITY TUNER. SIZE ONLY 6in, x 4in, x 24in, 3 I.F. stages. Double tuned discriminator. Ample output to feed most amplifiers. Operates on 9 volt battery, Coverago 88-108 Mc/s. Ready built ready for use. Fantastic value for

OUR £5.95 P. 8 P. Stereo Multiplex Adaptor £4.97.

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SH628 STERFO **HEADPHONES**



Outstanding value. Soft earpads, adjustable headband. 8-16 ohm s. 20-20,000 Hz. Complete with lead and stereo plug.

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£1.87 30p

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Sensitive magnetic headset with soft earpads. Impedance 2,600 ohms (d.c. 600 ohms). Frequency response 200-4000 Hz.

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Lightweight head-phones with padded earpieces. 4-16 ohms. 20-20,000 Hz. Com-plete with 6ft cord and plug,

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SDH8V MONO/STEREO **HEADPHONES**



Two way stereo/mono with vol-ume controls.
Padded head-heand. 4-16 ohms.
20-18,000 Hz.
Complete with
lead and stereo
plug.

OUR P. & P. £4.97 PRICE 30p

BH001 HEADSET AND BOOM MICROPHONE



Moving coil. Headphone imp. 16 ohms. Mike imp. 200 ohms. Ideal for lan-guage teaching, communications etc. Complete with leads and plugs.

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DH.08S Stereo Headphones



De luxe model with unique 2 way mechanical units and volume controls. 8 o h m s. 2 0 - 20,000 Hz. Complete with coil lead and stereo jack plug.

P. & P.

OUR 30p PRICE



Soft leather ear cushions. Impedance 8-16 ohms. 20-20,000 Hz. 15ft. coiled lead. Brand New and boxed. (List 69.95).

£6·75 30p PRICE **4-CHANNEL STEREO**

HEADPHONES TTC G3600
—Soft vinyl
covered head
cushion and
earphones.
Each earplece incorporates two
2" speaker
units. Fitted
2/4 channel
change over
switch. Impec
response 20-2



ce 4-16 ohm. Frequency response 20-20,000 Hz. Complete 15ft. coiled lead fitted two stereo reo plugs. P. & P. OUR £9.95

Audiotronic Products are manufactured exclusively for the Audiotronic Group of Companies and as member of the group we are pleased to offer you this fabulous range of high quality equipment. Made to our own specifications each item provides outstanding performance and reliability at a value for money price!

ACR.14 BATTERY/ MAINS CASSETTE RECORDER

Portable twin track mono recorder with automatic recording level control. Built-in speaker. Earpiece socket. Input for radio or record player. Fast forward and rewind. Output 500mW. AC 220/240v. or 6v. DC operation. Complete with remote control microphone, mains lead, earpiece and batteries.



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AHP-8D 8 TRACK STEREO TAPE DECK

Can be used with most hi-fi amplifiers. Push button track selector and illustrated track indicators. Attractive cabinet with black and silver trim. Output level 750mV. AC 220/240v.



UUR **£11-95** P. & P. 50p





£5.99 £8.59 £10.63 Post

MULTIBAND RADIO



ONLY £7.50 PAP35P

SQ. QUADRAPHONIC **DECODER**



Fully transistorised containing all input, output and control facilities to reproduce 4 channel sound from 80 records and 4 channel discrete, sources. Also provides conventional stereo corproduction, 2-channel and 4-channel tape recording and playback and enhancement of conventional 2-channel material. AC 220-240v. Frequency response ± 1dBm. 6Hz to 100KHz. Size: 180mm × 55mm × 220mm deep.

AHP-8A 8 TRACK STEREO TAPE PLAYER

Incorporates built-in amplifiers giving 44+44 watts rms output. Push butten track selector, illuminated track indicators, sider controls for volume, balance and tone. Attractive calinet with black and silver trim. Output impedance 8 ohms. AC 220/240v.



PRICE £17.25 P. & P. 500



AUDIOTRONIC DOLBY 'B' NOISE REDUCTION UNITS

Reduce tape hiss by 3dB at 600Hz, 6dB at 1200Hz and 10dB for all frequencies above 3000Hz. Size 16% × 8" × 3%. AC 200/250v.

PROCESS TWO

For use with cassette and tape recorders. Freq. res. 30Hz-20KHz+26B. Off tape monitoring. Switchable multiplex filter. Two Dolby calibration meters. S/N better than 70dB. Supplied with test cassette or tape as required.

OUR FRICE £34.50 P. & P.

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For use with semi professional tape recorders. Freq. res. 30Hz—20KHz±2dB. SN better than 70dB. Full source tape monitoring. Record/Replay metering, Switchable multiplex filter. Supplied with text table.

OUR £50.00 P. & P.

STEREO HEADPHONES

LSH.20 Individual volume controls. Stereo mono switch. 8 ohms, 40-8 ohms. 40-19,000 Hz. **£3.50.** P & P 30p.



LSH.40 Two way speaker system. nIdividual volume controls. 8 ohms. LSH.40 Two 20-20,000 Hz. **£6-95.** P & P 30p.



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LSH.50 Professional Quality Electro-static. Complete with self powered energiser and con-trol unit with headphone/speaker sel-ector. 4-32 ohms. 20-24,000 Hz. £15-95. P & P 30p.

SPORTSMAN AM/FM **PORTABLE RADIO** MODEL AR.1000

INJUEL AN. 1000

5 wavebands covering
AM 535-1065KHz. PM
88-107MHz. AIR 108135MHz. PB 147174MHz. WR 162.5MHz.
Large horizontal slide
dist with logging scale.
Slider volume and
equelch controls. 7 section telescopic aerial for
PM and built-in ferrite bar for AM.
APC. 3In. speaker. Earplece socket.
Green leatherette covered cabinet with
metal side panels. Size 152 x 79 x
215mm. Batterylmains operation.

OIIR C44.ET. P. S. P.

£11.50 P. 8 P. OHR PRICE



17+17 watt amplifier, Garrard AP76, plinth and cover, G800 cartridge, pair of Wharfdale Linton 2 speakers and all

OUR **£86-95** Carr. & Ins. OUR

Matching LT1700 AM/FM Stereo Tuner £37.50.



12v neg. earth. Slider controls for Volume, Tone and Balance. Channel selector button with red pilot lamp. Complete with speakers, mounting brackets and instructions.

ONLY £12.50 P&P



ACR 3500 CAR RADIO

Manual tuning of Medium and Long waves. 12v pos. or neg. earth. Complete with speaker, mounting brackets and instructions.

OUR **£6.50** P& P



CAR RADIO Push button tuning of one LW and five MW stations of

earth. Complete with spea-ker, mounting brackets and instructions. OUR E8-95 PAP



5 section. Fully au-tomatic. 12v DC. Extends to approx. 40". Complete with switch, all leads and instructions.

OUR PRICE £5.95

P&P 50p

TO 33%% OR M



AAG300 AM/FM STEREO TUNER AMPLIFIER

20+20 watts rms. Magnetic, ceramic and tape inputs. FM 88-108 MHz. AM 535-1605 kHz. Dual stereo speaker outputs. Headphone socket. (Rec. List Price £117.46).

| OUR | CCI | l-95 | P. & I | ρ. |
|----------------------------------|------------------|---|-------------------|----------------|
| PRICE | LU | נכיו | 75 | D |
| | | | | • |
| CASSETT: CS35D De | | '. 50p) | £44.4 | ıĸ |
| CS35 Reco | ck | •• | £57.8 | |
| CS35/CSS8 | 3 Speaker | 8 | \$67.8 |) 5 |
| GXC40D | Deck | | £57 £ | |
| GXC40 R | ecorder | deron. | £67.2 | |
| GXC45 D | eck | | £78-2 | |
| GXC46D | Dollby De | ck | £83·5 | 55 |
| GXC46 R | ecorder | | £89.9 | |
| GXC60D GXC65D CARTRID | Deck Dolby De | olr . | £87·6 | |
| CARTRID | GE (P. & | P. 50p) | 202 | ,,, |
| CR8I Dec CR81D De CR81T Re | k with an | ips. | £65·4 | |
| CR81D D | eck | •.* | £53·6 | |
| CR81T Re | corder/Ke | eceiver Pagardor | £114.2 | |
| CR80D88 | 4 channe | el Record | ier £93 .6 | |
| CR80DSS TAPE (P. | & P. 75p |) | | |
| 4000DS D | eck | | £59⋅9 | |
| 4000DS D | ust Cover | ` :: | \$3.9 |)0)5 |
| 1721L Re- X5000 Re | coruer corder | • • • | £60.5 | អា |
| X201D D | eek | • | £108-2 | |
| GX220D | Deck | | #123 8 | |
| GX221D GX280D | D J. | • • | £138·4 | |
| GX280D | Deck ook | • • | £211-£ | |
| TAPE/CA | | P. & P. 7 | | • |
| GX1900D | | | £144-5 | 60 |
| TAPE/CA | | (P. & P | | |
| X1810D I | Deck | | £138·4 | 10 |
| TAPE/CA | SETTE/C | ARTRID | GE | |
| (P. & P. 7 X2000SD | Decorder | | £180-7 | 7 K |
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| 2192 | 0·40 0·40 | 2N3878 | 1.22 | AC154 | 0.20 | BC182L | 0.12 | BF166 | 0.35 | BSY54 0 | 0 INST73 (1.5 amp 4) | | CL1005 (10 amp 600 | pv) |
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| 2222 | 0.31 | 2N4060 2N4061 | 0·11 0·11 | ACY40 | 0.17 | BC212K | 0.10 | BF196 | 0.15 | GET115 0- | e IN914 7p B. | 142 17p | BYZ10 35p 0A81 | |
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| 2647 | 1.20 | 2N4914 2N4915 | 0·87 0·95 | AD143 | 0.45 | BC239 | 0.09 | BF225J | 0.19 | GET873 0- | | /126 15p /127 17½p | 0A47 7½p 0A20 0A70 7½p 0A20 | |
| 2711 2712 | 0·12 0·12 | 2N4916 | 0.20 | AD149V AD150 | 0·66 0·63 | BC251 BC252 | 0·20 0·18 | BF237 BF238 | 0·22 0·22 | GET880 0- GET883 0- | | /140 £1.00 | 0A70 720 0A20 | ,,, |
| 2713 | 0·17 0·17 | 2N4917 2N4918 | 0·17 0·50 | AD161 AD162 | 0·49 0·51 | BC253 BC257 | 0·23 0·09 | BF244 BF245 | 0·16 0·33 | GET887 0- GET890 0- | 0 | | POTENTIOMETERS | |
| 2714 2904 | 0.18 | 2N4919 | 0.53 | AD161 \ | Pr. | BC258 | 0.09 | BF246 | 0.43 | GET895 0 | 5 MINITRON 3015F | 7-SEGMENT | Carbon: | |
| 2904A 2905 | 0·25 0·26 | 2N4920 2N4921 | 0·60 0·50 | AD162 ∫ ADZ11 | 1·15 1·50 | BC259 BC261 | 0·13 0·20 | BF247 BF254 | 0·49 0·14 | TIP29A 0- TIP30A 0- | _ 110010711011 (10 111 | | Log. or Lin., less swi | |
| 2905A | 0.23 | 2N4922 | 0·55 0·60 | ADZ12 | 1.75 | BC262 | 0.18 | BF255 | 0.15 | TIP31A 0- | 2 COCKETS | £1·30 20p | Log. or Lin., with sw Wire-wound Pots (3) | |
| 2906 2906 A | 0·18 0·23 | 2N4923 2N4926 | 0.80 | AF106 AF109R | 0·27 0·40 | BC263 BC300 | 0·23 0·42 | BF257 BF258 | 0·41 0·46 | TIP32A 0- TIP33A 1- | 4 | EMITTING | Twin Ganged Stereo | |
| 2907 | 0.18 | 2N4927 2N4928 | 1·00 1·80 | AF114 | 0.25 | BC301 | 0∙34 | BF259 | 0.48 | TIP34A 1. | 1 DIODE. (Red). 35 | | or Lin., 43 p | |
| 2907A 2923 | 0·25 0·12 | 2N4929 | 2.23 | AF115 AF116 | 0·24 0·25 | BC302 BC303 | 0·27 0·54 | BF261 BF264 | 0·23 1·45 | TIP35A 2- TIP36A 3- | Baann SHOINKES | | PRESETS (CARBON | |
| 2924 | 0·12 | 2N4930 2N4931 | 2·25 2·70 | AF117 | 6.20 | BC304 | 0.43 | BF270 | 0.25 | TIP41A 0- | 9 SCORPIO ignition | kit £10 + | | RTICA Or |
| 2925 2926 | 0 ·12 | 2N5172 | 0.12 | AF118 AF121 | 0·50 0·22 | BC307 BC307A | 0·10 0·10 | BF271 BF272 | 0·21 0·53 | TIP42A 0- | 8 50,11411 | | 0.3 Watt 7½p HORIZ | |
| Freen | 0.10 | 2N5174 | 0.22 | AF124 | 0.24 | BC307VI | 0.10 | BF273 | 0.25 | TIP3055 0- | io WIRE-WOUND R | | SLIDE POTENTIOM | |
| rellow Orange | 0·10 0·10 | 2N5175 2N5176 | 0·26 0·32 | AF125 AF126 | 0·20 0·19 | BC308 BC308A | 0·09 | BF274 BF457 | 0·28 0·53 | | 2.5 watt 5% (up t only), 7p | 0 2/0 DIMS | 58mm. TRAC | |
| 3053 | 0·31 0·46 | 2N5190 2N5191 | 0·92 0·96 | AF127 | 0·20 0·38 | BC308B | 0·09 0·10 | BF458 | 6-65 | RETURN | 5 watt 5 + (up to 8 | ·2kΩ only), | STNGLE GANGED, LO |)G or |
| | 0.46 | 2N5192 | 1.24 | AF139 AF170 | 0.25 | BC309 BC309A | 0.10 | BF459 BFS21 | 0·57 2·10 | OF | 9p | 2610 001-1 | 1k to 1M. 40p | |
| 3054 | | 2N5193 | 1.01 | AF172 | 0.25 | BC309B BC313 | 0·10 0·30 | BFS21A | 2·30 0·92 | POST | 10 watt 5% (up to 10p | ZUK32 ONIY), | TWIN GANGED, LOG 1k to 500k. 60p | |
| 3054 3055 3390 | 0.20 | | 4.40 | A 5470 | | | 0.30 | BFS28 | 0.92 | | 1 .04 | 1 | to obout coh | |
| 3054 3055 3390 3391 3391 A | 0·20 0·22 | 2N5194 2N5195 | 1·10 1·46 | AF178 AF179 | 0·55 0·65 | BC327 | 0.24 | BFS61 | 0.27 | | | CMALL | VALUE | |
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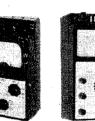


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Advertisements accepted up to 12 p.m., Thursday, July 5th for the August issue, subject to space being available.

Capital Radio

Operations and Engineering Staff

Capital Radio, having been awarded the contract for the London General Radio Station by the IBA, is now seeking talented engineers and operations staff for their studio in Central London.

Capital Radio will be the most important independent entertainment radio station in the United Kingdom and the staff we are looking for should be capable and willing to meet the challenge this involves.

The following posts are open:

- 1. ASSISTANT CHIEF ENGINEER Applicants should be fully conversant with modern broadcasting technology and capable of giving a lead to a team of first class maintenance and development engineers.
- 2. SOUND SUPERVISOR This is a key position in the operation of Capital Radio. The man we seek should have had a wide experience of operational sound working, either in radio, television or the recording industry. He should be capable of multi-channel operations in stereo and will be responsible to the Chief Engineer for all operational duties.
 - Salary and conditions for both these appointments will reflect the importance we place on them.
- 3. ENGINEERS We also require first class engineers for operations, maintenance and development work on our modern and sophisticated studio equipment. Experience in a broadcasting environment is necessary.

Salary range: £1,992 — £2,944

4. SOUND TECHNICIANS — Capable of operation duties in a highly demanding area. A youthful outlook and enthusiastic interest in developing programme techniques is essential. Operational experience will be necessary.

Salary range: £1,992 — £2,833

Applicants for all these posts should submit full details of their education, experience and present salary to:—

The Chief Engineer, Capital Radio Limited, 96 Piccadilly, London W.1.

2809



With the expansion of the company's offshore oil and gas production complex within the UK and Arabian Gulf areas. a telecommunications engineer is required to work in our London Office on the design, development and implementation of microwave and ultra high frequency radio systems including tropospheric scatter and digital remote control and supervisory systems. This includes interfacing with measurement and controlling instruments, pipeline valves and pumps, etc., for remote offshore oil and gas field production systems and pipelines and telecommunications systems for normal correspondence and date.

Duties will include initial site surveys and supervising final stage installation and commissioning involving overseas travel. Planning and engagement of local engineering staff for new systems and liaison with these engineers thereafter will be an important part of the duties.

Candidates aged 30 to 40 must be qualified C.Eng. or B.Sc Electronic Engineering. Experience of at least four years' in a senior capacity in installation testing or site maintenance of electronic equipment and four years' design and development experience in systems using this equipment is essential. At least one year's experience in the oil or allied industries is desirable.

■ Please write, quoting reference ZH341, giving age and brief details of background and previous experience to: The Manager, Central Recruitment, The British Petroleum Company Limited, Britannic House, Moor Lane, London, EC2Y 9BU.

HER MAJESTY'S GOVERNMENT COMMUNICATIONS CENTRE HANSLOPE PARK, MILTON KEYNES, MK19 7BH

has vacancies in the following fields of work

- (a) Microwaves.
- (b) HF Communications.
- (c) VHF/UHF Communications.
- (d) Acoustics.

- (e) General Electronic Circuit Design.
- Operational Analysis. For these posts applicants should be experienced scientists/engineers who have moved into Operational Analysis rather than the inverse.

Most of the posts are at Hanslope Park but some will be in the London Area.

Appointments will be made within the grades of Scientific Officer, Higher Scientific Officer and Senior Scientific Officer in accordance with the following definitions:

SCIENTIFIC OFFICER

Applicants should be not more than 27 years of age and should have one of the following qualifications:

- A degree in a scientific or engineering subject.
- Degree-standard membership of a Professional Institution.
- higher National Certificate or Higher National Diploma in a scientific or engineering subject.
- A qualification equivalent to (c) above. Salary Scales: £1206-£2043 with the entry point determined by qualifications and experience.

HIGHER SCIENTIFIC OFFICER

Applicants should be under 30 years of age but this requirement may be waived if special qualifications or experience can be offered. Formal qualifications are the same as for Scientific Officer above but in addition the following experience is required:

Applicants with 1st or 2nd class honours degrees—at least 2 years postgraduate experience.

Applicants with other qualifications—at least 5 years post qualification experience.

Salary Scale: £1946-£2515 with entry point dependent upon experience.

Applications stating the field of work and grade required should be made to:

value can be offered

A 1st or 2nd class honours degree with at least 4 years post-graduate experience is the normal requirement for this grade although

SENIOR

SCIENTIFIC OFFICER

Applicants should be at least 25 and under

years of age, although the upper age limit may be waived if experience of special

applicants with the other qualifications given above may be considered if they have had at least 7 years appropriate experience. Salary Scale: £2464-£3483. Entry will normally be at the minimum of the scale

but applicants with experience of special value may be entered above the minimum.

ADMINISTRATION OFFICER,

H.M. GOVERNMENT COMMUNICATIONS CENTRE, HANSLOPE PARK, HANSLOPE, MILTON KEYNES, MK19 7BH

RADIO OFFICERS would you come ashore for £2,300 a year?

As a Radio Operator with the Post Office Maritime Service you can continue your career ashore in an interesting and expanding service. And earn over £2,000 a year, including compulsory pension contributions, at 25 years of age working only a 41-hour week of shift duties—with overtime this could rise to £2,300 and possibly more.

Post Office Radio Operators benefit from a shorter pay scale than sea-going officers. You have good opportunities for promotion to positions earning basic salaries of up to £3,290, and prospects of further advancement into Post Office Senior

Management.

To apply you need to be 21 or over and to hold a 1st class or General Certificate issued by the MPT or an equivalent certificate issued by a Commonwealth administration or the Irish Republic.

If you would like to know more, please write to the Inspector of Wireless Telegraphy, Post Office, IMTR/WTS1.1.3, Union House, St. Martin's-le-Grand, London EC1A 1AR. L49

Post Office Telecommunications

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SPANISH COMMUNICATIONS EQUIPMENT MANUFACTURER

Applications are invited from qualified design engineers specialized on:

- a) Ground/Air Communications
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At least 5 years experience desirable. Company located in Madrid. Salary open.

Send resumé to:

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Fernando el Católico, 63 Madrid 15 SPAIN

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Granada Television

Electronic Engineers for Operational Television

Expansion has created further vacancies at the TV Centre in Manchester for electronic engineers to work in all aspects of Granada's broadcasting operations, covering studio vision, videotape, telecine, transmission switching and maintenance of sound and vision equipment.

Vacancies exist at all levels from 1st year Engineer to Substantive Grade (6th year) Engineer at salaries between £1992 and £2944 per annum. Entry point will depend on experience. Candidates who lack experience in broadcast engineering will need a minimum of two years experience in communications engineering and preferably an ONC or equivalent. Shift work, including weekends. Generous Granada Group pension and life assurance benefits. Four weeks holiday.

Applications, stating full details of education and experience to:

Robert Connell Granada Television MANCHESTER M60 9EA

GRANADA



Audio Frequency Circuit Designer

Advanced design work now being undertaken at STL has world-wide implications, both in the more conventional areas of subscriber telephones and in new electro-acoustic devices and display systems. Activity is concentrated on analogue and digital, bipolar and MOS integrated circuits at audio frequencies

We need a designer to join our telephone subset component division, working on audio frequency circuits. He will be a graduate with at least three years' experience in circuit design. Phases of design vary according to individual problems, but in general he will work, after on-the-job training, in feasibility studies, preparation of design and procurement specifications, breadboarding and circuit optimisation constructing prototypes, liaising with the drawing office to final engineering drawings, and following jobs through to production.

Salary and career prospects, both in the UK and with associate companies in Europe, are excellent.

Location is a modern, well-equipped laboratory complex in pleasant rural surroundings. Help will be given with relocation expenses.

This is one of a number of key appointments we are making because of the expansion of our activities. A team of our engineers and managers is touring the country for informal interviews with engineers and managers of many disciplines. They will be pleased to meet you and talk about your prospects with STL, in the UK and Europe.

Come and see us, between 4pm and 10pm (or the morning after between 8.30am and 10am) at:

JUNE 14th St. Albans – The Great Red Lion Hotel
JUNE 19th Chelmsford – County Hotel
JUNE 21st Birmingham – Albany Hotel, Smallbrook
JUNE 22nd Bournemouth – Pavilion Hotel, Bath Road
JUNE 27th London – Waldorf Hotel, Aldwych, WC2

If you can't make these dates, please telephone or write to: Martin Jenner,
Standard Telecommunication Laboratories Limited,

London Road, Harlow, Essex. Tel: Harlow 29531.

Standard Telecommunication Laboratories

A British Research Centre of ITT

In-Car Entertainment SALES DEVELOPMENT MANAGER

A major distributive company growing nationally and diversifying the range of manufacturers products and own company services supplied to the retail and commercial trades is developing through its many wholesale distribution outlets the sales of a complete range of IN-CAR ENTERTAINMENT equipment. An installation service is also to be provided.

This is a new venture in a company which has previously had only minor involvement in this field.

We want to recruit a man with some technical knowledge of the ICE products range, relevant sales and marketing skills and the ability to lift this project off the ground and establish our ICE activities as a profitable and viable proposition.

This post will be of interest to men now earning around £2,500 p.a. Company car and other benefits associated with a big company apply.

Write in strict confidence, giving full career history and personal details to date.

Please advise in a separate letter, any company to whom you do not wish your application to be forwarded.

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Pike Russell and Associates, Ref 166, St. Vedast House, 150 Cheapside, London EC2

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ELECTRONIC DESK CALCULATORS PROGRAMMABLE CALCULATORS VISIBLE RECORD COMPUTORS PERIPHERALS

There are vacancies in our Field Service Organisation for Engineers to service the above range of equipment installed in London and the Home Counties.

Applications are invited from :-

- Electronic Engineers qualified to Intermediate City & Guilds Certificate or equivalent standard and
- Electro/Mechanical Engineers experienced in Triumph/Adler and/or IBM input/output typewriters, readers and punches.

Excellent training facilities and first class conditions of employment.

For further information please contact:-

Administration Manager, Sumlock Comptometer Ltd., Anita House, Rockingham Road, Uxbridge, Middlesex. Tel: 89-51522





Audio, Digital & Telephone Engineers...

3-way potential at Cambridge

Several rare, possibly unique, positions concerning design of audio distribution systems and control logic, with involvement in digital, audio and telephone engineering all at the same time. In the compact, highly professional Advanced Systems Department, the unit which has contributed heavily to the Company's firmly-established international leadership in radio communications.

Your experience should preferably span all three fields involved, though it may currently be much deeper in one than the others. Practical ability, together with a commercially-minded attitude are more important than qualifications, although these would naturally be advantageous. You must be strong in logic as well as innovation, with the ability not simply to conceive preliminary design ideas but to visualise the end result at the same time.

For more information, including salaries, your potential career within this extensive Group, and Company benefits such as relocation assistance, please give a summary of your background in your letter addressed to M. W. Timmins, Senior Personnel Officer,



ROBERT GORDON'S INSTITUTE OF TECHNOLOGY, ABERDEEN SCHOOL OF ELECTRONIC AND ELECTRICAL **ENGINEERING**

ELECTRONICS WORKSHOP TECHNICIAN

Applicants for the above post should have appropriate qualifications and several years of experience in the field of linear and digital electronic circuits and systems.

Salary on the scale £1,728 to £2,241 per annum with placing determined by qualifications and experience.

Further particulars and forms of application may be obtained from:-

The Secretary and Treasurer Robert Gordon's Institute of Technology ABERDEEN, AB9 1FR

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UNIVERSITY OF SURREY TELEVISION SERVICE

SENIOR ENGINEE

(Technician Grade 5 £1,881—£2,241)

The Television Service is an integral part of the Audio Visual Aids Unit which provides services for teaching and research illustration throughout the University. The television equipment includes a most versatile mobile control room handling three Plumbicon studio cameras and five others.

The Senior Engineer will be responsible to the Chief Engineer/Operations Manager and will assist with the operation and maintenance of all CCTV equipment including IVC video-tape recorders and Philips Plumbicon cameras.

The minimum qualification is HNC in electronics or electrical engineering or an appropriate training in the BBC or a programme company. Experience in educational television is desirable.

Applications are invited immediately on forms which may be obtained from the Assistant Secretary (Personnel), University of Surrey, Guildford, Surrey, Previous applicants need not apply. Tel: Guildford 71281 Ext. 452.
Closing date: 6 July 1973.

12787

ANTARCTIC **EXPEDITION**

Vacancies exist for Ionospherics Technicians to work for about 30 months in the Antarctic under expedition conditions. Candidates, preferably single and aged 21-30, must be physically fit, qualified in electronics and possessing a good groundwork in valves techniques. They should be capable of working without supervision and able to improvise in adverse situations.

Successful applicants will be given a training course prior to departure in October and will be expected to commence this immediately.

Salary from £1,442 with annual increments and low income tax. Clothing provided and free messing aboard ship and in Antarctica.

Applications to:

BRITISH ANTARCTIC SURVEY. 30 Gillingham Street, London, S.W.1 Tel: (01)-834 3687

Project and **Commissioning Engineer**

Should have good theoretical training in electronics plus actual experience of professional audio equipment in broadcasting, recording, or with a specialist manufacturer.

Electronics Draughtsman

To work mainly on console structures, chassis and panel layouts, (small electro-mechanical details), and some printed circuit work. At least two years similar experience. Apply in writing to:

Mr. Swettenham, HELIOS ELECTRONICS LIMITED, 161 High Street, Teddington, Middlesex, TW11 8HT

CHELSEA COLLEGE University of London

Electronics Technician Grade V

Allowance).

required in Applied Acoustics Laboratories for the design, development and maintenance of electronic systems for postgraduate teaching and research. Salary Scale £2,056-£2,416 per annum (including London

Further details and application forms

The Departmental Superintendent (5AA), Chelsea College, (WW) Pulton Place, London, SW6 5PR.

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UNIVERSITY OF SOUTHAMPTON

Department of Physics

TECHNICIAN

is required to assist with the construction and assembly of instrumentation to be flown in small sounding rockets for upper atmosphere research.

Applicants should have served an apprenticeship or hold an equivalent qualification in electrical engineering with relevant workshop experience. Additional experience in electronics would be an advantage.

an advantage.

The appointment will be for one year in the first instance. Salary scale £1,539—£1,794 per annum. Applications giving details of age, qualifications and experience and the names and addresses of two referees should be sent to the Deputy Secretary's Section (Ext. 2400), The University, Southampton SO9 5NH, as soon as possible. Please quote reference number: WW 259/73/T.

FOREMOST IN THE UNITED KINGDOM

IN CONSTANT TOUCH WITH EVERY EMPLOYER OF EXPERIENCED ELECTRONICS **ENGINEERS**

Our professional placement service is specialised, confidential and completely

Phone us at any time or write quoting WW105

ELECTRONICS APPOINTMENTS LTD 4 DRYDEN CHAMBERS. 119 OXFORD STREET LONDON, W1R 1PA.

MARCONI INSTRUMENTS LIMITED

ELECTRONIC CHNICIA

are required to work on calibration, fault-finding and testing of telecommunications measuring instruments. The work is varied and will enable technicians with experience of r.f. circuits to broaden their knowledge of the latest techniques employed in the electronics and telecommunications industries by bringing them into contact with a wide range of the most advanced measuring instruments embracing all frequencies up to u.h.f.

Entrants may be graded as Test Technicians, Senior Test Technicians or Technician Engineers according to experience and qualifications. Our servicing and production programme, geared to our recognised export achievement, provides employment combined with prospects of advancement, not only within these grades, but into other technical and supervisory posts within the Company at Luton and St. Albans.

Salaries are attractive and conditions excellent. A Pension Scheme includes substantial life assurance cover provided by the Company. Assistance with removal may also be given in appropriate cases. Please write or telephone, quoting reference WW178 for application form to:



Mr. M. Leavens, Works Manager Telephone: Luton 33866, or Mr P Elsip, Personnel Officer Marconi Instruments Ltd Longacres, St. Albans, Herts Telephone: St. Albans 59292



Member of GEC-Marconi Electronics

LOUDSPEAKER DESIGN ENGINEER

An outstanding opportunity exists for an enthusiastic engineer to set up a new division of Loudspeaker Manufacturing within the SNS Electronics Group

Experience in design and production management techniques in the Audio Electronics Industry is required, together with a determination

The successful applicant, who will be responsible to the Technical Director, will join a young and expanding organisation with a reputation for equipment of the highest quality, in which prospects for advancement are excellent.

Salary will be commensurate with ability, experience and qualifications.

Telephone or write for an appointment to:-

Mr. R. Jones, Technical Director, **SNS ELECTRONICS GROUP** 851 Ringwood Road, Bournemouth, BH11 8LN Tel: Northbourne (02016) 5331/4.



Careers with EMI in

Quality Assurance

A rapid expansion programme throughout our location at Hayes, Middlesex, has created vacancies for Test and Quality Staff.

Senior Test Engineers Test Engineers Assistant Test Engineers

If you are interested in working in any of the areas outlined below, please contact us. We are one of the world's leading companies in electronics and we can offer you a wide range of opportunities and excellent prospects for a satisfying career in engineering.

Radar and Equipment Division

This division designs and manufactures sophisticated airborn and surface radar equipment. We have a number of large contracts for the development and supply of radar devices to the armed forces.

Sound and Vision Equipment Ltd.

This company designs and manufactures audio-visual equipment including closed-circuit television systems, television cameras and video devices. It won the Queens Award to Industry for its celebrated 2000 range of light-weight television cameras.

Medical Electronics Division

This division designs and produces the celebrated EMI-Scanner. The computerised X-ray system for the diagnosis of brain diseases, invented by one of our research scientists who won the £25,000 MacRobert Award for outstanding technological innovation in British Industry in 1971. This equipment obtained the 1973 Queen's Award to Industry for technological achievement.

Microelectronics Division

The work of this division is the development and manufacture of thick and thin film circuits and a wide range of microelectronic devices.

Central Quality Assurance

We have the following vacancies in our Central Quality Assurance Department which provides technical support to all divisions in the electronics operations.

Test Gear Engineer

To design and build special-to-type test equipment and to construct the calibration and maintenance schedules of test gear at the commissioning stage.

Technical Liaison Engineer

To liaise with development teams to provide transfer of expertise to the production units, advising on test methods and equipment and solving problems at all stages of development and production.

Metrologist

To carry out precision measurements in our standards laboratory using measuring devices, sensitive comparators and optical instruments.

Salaries: £1400-£2400

dependent upon the position and the experience and ability of the person Please write or telephone for an application form:



R. N. L. BLACK, Personnel Department, EMI Electronics Ltd., 135 Blyth Road, Hayes, Middlesex. Tel: 01-573 3888 Ext. 2887

International leaders in Electronics, Records and Entertainment.

DUBLIN INSTITUTE FOR ADVANCED STUDIES

School of Cosmic Physics

Applications are invited for the post of

SENIOR TECHNICIAN

in the Geophysical Section of the School located in Dublin. The necessary qualifications are the Science Laboratory Technicians Advanced Certificate, a University Degree in science or equivalents. The appointee will be required to maintain and operate laboratory and field geophysical equipment. Experience in maintenance, use and construction of

ELECTRONIC EQUIPMENT

is essential. The post is permanent and pensionable (non-contributory) with salary scale £1.675 \times £73 (5) \times £46—£2,086 (under review) plus Marriage and Family Allowances. Applications should reach the Registrar, Dublin Institute for Advanced Studies, 10 Burlington Road, Dublin 4, Ireland, not later than 31 July, 1973.

[2806

LINCOLN No. 1 HOSPITAL MANAGEMENT COMMITTEE

LINCOLNSHIRE MEDICAL PHYSICS DEPT.

MEDICAL PHYSICS TECHNICIAN

(GRADE III)

A technician is required to work primarily in the new Audiological Measurement section of the Lincolnshire Medical Physics Department at St. George's Hospital, Lincoln. The successful applicant will be required to assist in all the scientific audiological tests carried out by the section.

The work involves technical problems together with personal contact with patients of all ages.

Applicants should be qualified to HNC standard in Electronics or be trained to an equivalent standard. Comprehensive training will be given on technical and clinical practice. Further information from the Physicist-in-Charge.

Salary Scale £1,602—£2,076.

Applications giving full details of education, qualifications and experience together with the names of two referees to the Hospital Secretary, St. George's Hospital, Lincoln.

[278

HARROW COLLEGE
OF TECHNOLOGY AND ART

TECHNICIAN:

SCHOOL OF PHOTOGRAPHY

Up to £1,980

A Technician/Engineer in television and sound is required for a new post in the Film and Television Department. Duties will include the technical operation and routine maintenance of a C.C.T.V. Studio requiring electronic and mechanical skills and knowledge in the television field. Relevant City and Guilds or H.N.C. qualifications are desirable and further in-service training will be considered.

Unless otherwise stated all the above positions involve a 35-hour week (Monday to Friday).

Application forms from The Registrar, Harrow

Application forms from The Registrar, Harrow College of Technology and Art, Northwick Park, Harrow, Middx. HAI 3TP (864 4411 Ext. 31). Returnable by 8th June, 1973.

[2788

UNIVERSITY OF SURREY

Laboratory **Superintendent**

Department of Physics

Applications are invited for a newly established post in the Department of Physics. The person appointed must possess extensive experience of high level experimental work in scientific or engineering laboratories and will take an active part in experimental work in the Depart-

The qualification required is a degree in a relevant subject, but an experienced candidate with H.N.C. and corporate membership of an appropriate institution or learned society could be considered. The appointment will be that of a Senior Experimental Officer in the University for which the salary scale is £3,192-£4,299.

Further particulars can be obtained from the Assistant Secretary (Personnel), University of Surrey, Guildford, Surrey GU2 5XH, and to whom applications in the form of a curriculum vitae should be returned to by 6th July, 1973.

UNIVERSITY OF KENT AT CANTERBURY

Research Studentships in Electronics

A few studentships are available for first and top second degree candidates for research in the following fields.

Digital Communications. Computer Systems and Cybernetics. Solid State Devices. Microwaves. Radio Astronomy. Space Research. Experimental Relativity.

Medical Electronics.

Further details and application forms from: Professor R. C. Jennison, The Electronics Laboratories, The University of Kent at Canterbury, Canterbury, Kent.

12792

ELECTRONICS TECHNICIAN

The Electronics Department of Brompton The Electronics Department of Brompton Hospital has a vacancy for an additional Electronics Technician (Grade 3) to undertake the construction and maintenance of electronic equipment used in the Hospital, e.g. instrumentation, amplifiers, transducers, oscilloscopes, optical recorders, mass-spectrometer, analogue computers and TV systems. Applicants should have a good working knowledge of electronics and be qualified to O.N.C. standard or equivalent. Day release facilities for further approved education.

Salary £1,714-£1,969 per annum including London Allowance. Further details and application form:

Departmental Superintendent, B.H., Chelsea College, Pulton Place, London SW6 5PR.

12692

Senior **lectronics**

Noise Measurement

We are seeking a man with a minimum qualification of an Electronics HNC to lead the Instrumentation Section of our Noise Research Department.

Based in Derby, this section provides a measurement service to a department of twenty-five qualified staff conducting research and development into methods of reducing aero-engine noise. Noise measurements are made on a variety of static rigs, engine test beds and on field exercises. The department is well equipped with modern instrumentation and excellent supporting facilities are available.

Candidates should be aged between 28 and 35 and be prepared to work away from base for short periods during field trials.

At least four years experience of electronic instrumentation and test work is required, together with a sound knowledge of analogue tape recording. Some experience of acoustic measurements and Bruel and Kjaer equipment would be advantageous.

The Company operates an excellent Life Assurance and Staff Pension Scheme, and we shall be pleased to discuss relocation assistance with applicants who are invited for interview.

Please write for an application form, giving brief details of qualifications and career to-date, to: Head of Employment Office,

Rolls-Royce (1971) Ltd, Derby Engine Division, P.O. Box 31, Derby DE2 8BJ

Derby Engine Division

BOTSWANA ASSISTANT ENGINEER Grade II

- ★ Gratuity 25% of basic salary.
- Subsidised accommodation.
- 24-36-month tour.
- Appointment Grant £100-£200 normally payable.
- Low taxation.
- Holiday visit passages. Education Allowances.
- * Free family passages.

Required by the Posts and Telecommunications Department to establish a new radio workshop repairing a wide range of HF/VHF radio communications

Candidates, 30-45 years, must possess the City and Guilds Intermediate Certificate in Telecommunications, have had five years' relevant experience, excluding training, and must be capable of undertaking minor radio equipment modification, installation and aerial work. Experience with Plessey, Racal and Pye equipment and some knowledge of radio stores would be an advantage.

Salary will be in the range £1780 to £2700 according to experience.

Because of lower rates of Income Tax in Botswana the gross emoluments, including gratuity, are roughly equivalent to a UK salary of £2600 to £3700 for a single man and £2650 to £3900 for a married man with two children.

This post is partly financed by Britain's programme of aid to the developing countries administered by the Overseas Development Administration of the Foreign and Commonwealth Office.

For further particulars you should apply, giving brief details of experience to:

M Division, 4 Millbank, London, SW1P 3JD, quoting reference number M2K/730429/WF.

Electronics

Pye Telecommunications of Cambridge and Haverhill have immediate vacancies for Production Test Engineers. The work entails checking to an exacting specification VHF/UHF radio-telephone equipment before customer delivery; applicants must therefore have experience of fault finding and testing electronic equipment, preferably communications equipment. Formal qualifications while desirable, are not as important as practical proficiency. Armed service experience of such work would be perfectly acceptable. Pye Telecommunications is the world's largest exporter of radio-telephone equipment and is engaged in a major expansion programme designed to double present turnover during the next five years. There are, therefore, excellent opportunities for promotion within the company. Pye also encourages its staff to take higher technical and professional qualifications.

These are genuine career opportunities in an expansionist company, so write or telephone without delay for an application form to:

Mrs A E Darkin at

Cambridge Works, Elizabeth Way, Cambridge CB4 1DW.

Telephone: Cambridge 51351.

or Mrs C Dawe at Colne Valley Road, Haverhill, Suffolk.

Telephone: Haverhill 4422.



Pve Telecommunications Ltd

2802

RADIO/AUDIO ENGINEER

(CONSUMER ELECTRONICS)



FOR NEW ZEALAND

A versatile man is required with some experience in all or most of the following aspects; Component approval, development of mechanical chassis and presentation items as well as circuit development, a commercial appreciation of cost methods, industrial processes such as surface finishing of various materials and plastic moulding.

Appropriate formal qualification would be an advantage but broad and proven experience is considered most important.

This position should appeal to a man who prefers responsibility over a wide field rather than confinement to a narrow section of development engineering.

All applications will be carefully considered and suitable applicants will be interviewed by Thorn Consumer Electronics Ltd., formerly British Radio Corporation Ltd., Enfield, Middlesex.

The man finally chosen will travel to New Zealand by air under the New Zealand assisted immigration scheme for which all arrangements will be made by us.

Applicants should write giving full details of experience, marital status etc.. to:

THORN RADIO INDUSTRIES N.Z. LTD., 36/38 HASTIE AVE., MANGERS, AUCKLAND, NEW ZEALAND. a subsidiary of Thorn Consumer Electronics Ltd., England.

BRUNEL UNIVERSITY

COMPUTER SCIENCE

An

ELECTRONICS TECHNICIAN

is required from I August 1973 struction, modification, maintenance and minor development of Digital and Analogue equip-

This is a new post; experience of Digital or Analogue apparatus is desirable but not essential but a good grounding in electronics with some experience of service work is

Applicants preferably over 30, should possess a minimum of ONC or equivalent and/or substantial relevant experience.

Salary in the grade 4 scale, £1,728-£2,028 plus £90 London Weighting.

Postcard for application form and further details to Establishment Secretary, Brunel University, Uxbridge, Middlesex or telephone UXBRIDGE 37188, extension 40 (quoting reference CS/7). Closing date: 2nd July 1973.

Workshop Service Engineers

to repair calculator printed circuit boards. Good basic electronic knowledge required and experience in a Service Department. Salary £1,750 plus and L.V.'s.

Apply to:

Mr. V. Knight,

Automatic Business Machines
Wyfold Road, Fulham, S.W.6.

[2695 Tel: 385 3311

THE CITY UNIVERSITY

ELECTRONICS TECHNICIAN |

(Grade 3)

required immediately in the Educational Services Unit. It is expected he will have had experience in all forms of audio visual, audio recording, sound amplification and CCTV equipment, including video tape recording. Duties will include:

(a) Construction and modification of electronic equipment.

(a) Construction and modification of electronic equipment.
(b) Testing of equipment.
(c) Maintenance and adjustment of audio visual and CCTV equipment.
(d) Assist in maintaining a comprehensive audio visual service within the University.
(e) Ability to project film up to professional standard.
Salary within the range £1,539-£1,794 plus £175

Closing date: 29th June 1973.
Further details and application form can be obtained from The Personnel Officer, The City University, St. John Street, London, ECIV 498.

THE HATFIELD POLYTECHNIC

TECHNICIAN

for Psychological Laboratory

for maintenance and construction of a variety of electronic and other equipment.

Applicants should preferably hold an appropriate intermediate or National Certificate or City and Guilds quali-

Salary scale: £597-£1,143 per annum. Application form and further details from The Staffing Officer, The Hatfield Polytechnic, PO Box 109, Hatfield, Herts ALIO 9AB.

Please quote ref: 266/WW.

[2789



OPPORTUNITIES IN ADVANCED TECHNICAL TRAINING

APPLICATIONS ARE INVITED FROM

ELECTRONIC ENGINEERS OR TECHNICAL INSTRUCTORS WHO HAVE

- (a) A wide knowledge of electronics and computer techniques.
- (b) Experience in, and a flair for, trouble-shooting modern Digital hardware.
- (c) An interest in the education of young technicians and enthusiasm for transferring to them a significant amount of (a) and (b) above.

The positions will entail both formal instruction and convincing operational demonstrations before perceptive and conscientious students. An inventive and flexible approach will be reauired.

These advanced positions, which include some responsibility for the career development of students, will carry an attractive salary.

> Applications to the Personnel Department, DIGITAL EQUIPMENT INTERNATIONAL LTD., Industrual Estate, Galway, Ireland.



Wandsworth **Technical College**

Wandsworth High Street, London, SW18 2PP **TELEVISION STUDIO**

Senior Technician

up to £2,109

to be responsible for the general management of a large new well equipped Television Studio and Control Room.

He will be assisted by a Technician and will be responsible to the Vice-Principal. The Senior Technician will be required to become fully conversant with all the equipment and to assist users to operate it. Post Ref. 1ST.

Technician up to £1,788

Salary:
Senior Technician—on a scale within the range £1,629 - £2,019 with starting point and maximum dependant on qualifications and experience (plus £90 London Allowance). Technician—on a scale within the range £1,248 - £1,698 with starting point and maximum dependant on age, qualifications and experience (plus £90 London Allowance). Conditions include a basic 36½ hours per week with possibly some evening and overtime work.

week with possioly some evening and overtime work. Further details and application form from the Senior Administrative Officer: completed forms to be returned within 2 weeks of publication of this advertisement. Post Ref. 2T. [2659] 12659

NORTHWICK PARK HOSPITAL AND CLINICAL RESEARCH CENTRE Watford Road, Harrow, Middlesex, HA1 3UJ. Tel: 01-864 5311

ELECTRONICS TECHNICIANS

A Technician to service and calibrate a wide range of equipment used for medical, surgical and engineering purposes. The successful applicant will work closely with medical and other engineering purposes. professional staff.

Salary-Grade III: £1,728-£2,202, under review.

2. A Technician with duties similar to those described above who will be working under some supervision.

Salary-Grade IV: £1,548-£1,953, under review.

The hospital is new and closely allied with the Clinical Research Centre. In 1974/75 there will be over 900 beds. Good staff facilities and a pleasant working atmosphere. Active social club. Temporary accommodation may be available.

Further particulars and application forms, returnable by July 23rd, from Mrs. J. Moore, Ext. 2202.

RADIO OFFICERS

DO YOU HAVE

PMG 1 PMG 11, MPT 2 YEARS OPERATING EXPERIENCE. POSSESSION OF ONE OF THESE QUALIFIES YOU FOR CONSIDERATION FOR A RADIO OFFICER POST WITH COMPOSITE SIGNALS ORGANISATION.

On satisfactory completion af a 7 month specialist training course, successful applicants are paid on a scale rising to £2,527 pa; commencing salary according to age — 25 years and over £1807 pa. During training salary also by age, 25 and over £1350 pa with free accommodation.

The future holds good opportunities for established status, service overseas and promotion. Training courses commence at intervals throughout the year. Earliest possible application advised. Applications only from British-born UK residents up to 35 years of age (40 years if exceptionally well qualified) will be considered.

Full details from:

Recruitment Officer **Government Communications Headquarters** Room A/1105 Priors Road Oakley Cheltenham Glos GL52 5AJ

Telephone Cheltenham 21491 Ext 2270

Careers in Colour TV Systems and Radio Communications

If you have experience in the electronics industry, radio or T.V. maintenance or in the Services, we can offer you a progressive career in a Production Test Department here at Marconi Communication Systems in Chelmsford.

As a leading company in the field of Colour T.V. Systems, Line Transmission Equipment and Radio Communications, we can provide a wide variety of interesting work. There are excellent career prospects; many of our senior personnel started their careers with us in a Test Department.

You may be currently working in a dull, routine job in electronics, such as bulk production testing or domestic repair work, or you may feel that your future progress is limited. Even if you

doubt your suitability, come along and talk to us about your career development. Training will be available where necessary, together with day release facilities for young people.

Please write with full personal and career details to R.S. Ransom, Marconi Communication Systems Limited, New Street, Chelmsford, Essex, CM1 1PL, quoting ref WW or telephone Chelmsford 53221, extension 498.

2820



A GEC-Marconi Electronics Company

H.M. PRISON AND BORSTAL SERVICE

VOCATIONAL TRAINING INSTRUCTORS

(Civilian Instructional Officer, Grade III)

All applicants should have served a full apprenticeship or have had equivalent recognised training followed by at least five years industrial experience in the Radio and Television and/or Electronics servicing industry. City and Guilds Certificate (or equivalent) is desirable. Teaching or instructing experience are added advantages.

RADIO AND TELEVISION SERVICING posts at:
H.M. PRISON, AYLESBURY, Bucks.
H.M. PRISON, THE VERNE, Portland, Dorset.

Duties: The successful candidates will train inmates in Radio and Television servicing and prepare them for City and Guilds examinations.

Salary: The commencing salary at age 28 or over is £2,082 (slightly less if younger), rising to a maximum of £2,514. The posts are pensionable with prospects of establishment, i.e. permanent appointment.

Hours: A 40-hour, 5-day week is worked with 4 weeks and 2 days annual leave in addition to the usual 9 public and privilege holidays.

Write for Application Form to: The Establishment Officer, Home Office, Personnel and Administration Department, Portland House, R.10/10 (7TI), Stag Place, London, SW1E 5BX, stating for which post you apply. Closing date for the receipt of completed application forms: 29 June 1973.

2804

INSTITUTE OF GEOLOGICAL SCIENCES

GEOPHYSICAL INSTRUMENT TECHNICIAN

(P & TO IV)

required to assist in the maintenance, field trials, operation and development of complex geophysical instruments for use on land, at sea and in boreholes. Although based at Acton, field operations range over the British Isles and UK Continental Shelf and duties may also include short visits abroad.

Candidates must have ONC or equivalent in Electrical or Mechanical Engineering and have served a recognised apprenticeship.

Present salary £1,555 (at 21)—£1,940 (at 28 or over on entry)—£2,180. Subsistence is paid for field operations. Non-contributory pension scheme and prospects of promotion.

Application forms available from Establishment Section, Exhibition Road, London SW7 2DE. Telephone 01-589 3444. Please quote Ref. P & TO/AGU/73.

12697

SENIOR ELECTRONIC ENGINEER

required for circuit development. Experience of nanosecond pulse circuitry, design and familiarity with image converter devices would be an advantage. Several years' continuous experience in electronic circuit design is essential.

Qualifications of HND or HNC or equivalent required. Salary negotiable according to experience and qualifications.

We are a small specialist company manufacturing advanced scientific instruments. Very good working conditions in a pleasant rural area 25 minutes from Central London. Contributory pension and free-life insurance scheme. Excellent canteen facilities.

Apply:

JOHN HADLAND (P.I.) LTD., NEWHOUSE LABORATORIES, Bovingdon, Hemel Hempstead, Herts. Tel. Hemel Hempstead 832525.

[2816



DESIGN ENGINEER

20-30 year old General Electronic Design Engineer with some experience in Antenna design required for small expanding Company.

Must be capable of working by him-

He should be willing to assist on test supervision and occasional visits to customers.

Please apply to:

Mr. D. A. R. Wallace — Managing Director, Antenna Specialists UK Limited, Thame Industrial Estate, Bandet Way, Thame, Oxfordshire, Tel: Thame, 3621/2

BRITISH ANTARCTIC SURVEY

RADIO TECHNICIAN

required, three-year contract in Stanley, Falkland Islands. Qualified to ONC Tele-communications standard, experience of teleprinters advantageous. A knowledge of medium and high-powered HF transmitters and good general workshop ability with HF equipment. Officer will be required to carry out aerial farm maintenance as part of his duties. Salary at the rate of £1,379 per annum in the scale of £1,379, £1,443, £1,487, £1,570, £1,634, £1,745, £1,798. This scale is currently under revision.

Apply to:

BRITISH ANTARCTIC SURVEY, 30 Gillingham Street, London, S.W.1 Tel: (01)-834 3687

[2817

SAMUELSON FILM SERVICE LTD... LONDON, N.W.2 require an

ELECTRONIC ENGINEER

TO MAINTAIN PORTABLE RADIO TELEPHONE **EQUIPMENT**

Applicant must have a wide experience of fault finding and testing in this field.

Qualifications, while desirable. are not as important as practical proficiency.

> PENSION SCHEME SALARY NEGOTIABLE Telephone General Manager. 01-452 8090, Ext. 248

> > 12790

UNIVERSITY OF EDINBURGH **Department of Mechanical Engineering** Applications are invited for the post of

ELECTRONICS TECHNICIAN (TG5)

tenable from 1st August, 1973. The appointee will be expected to service the electronic equipment already in the Department, and to design and construct, using modern techniques, specialised equipment needed for research and teaching purposes. The possession of a H.N.C. may be an advantage.

Salary scale: £1,881-£2,241 per annum

Applications in writing, giving full details of experience and qualifications together with the names of two referees, should be sent, as soon as possible, to the Secretary to the University, Old College, South Bridge, Edinburgh, EH8 9YL. Please quote reference 2034.

ABILITY

IS THAT YOUR MIDDLE NAME?

Then come and join us at TELENG.

we need you

JUNIOR & SENIOR **ELECTRONIC** DEVELOPMENT **ENGINEERS**

are urgently required for our expanding UHF/VHF Television Distribution Laboratory. Write or phone for an immediate interview (which can be arranged outside normal working hours if necessary) to:

Mrs. V. NELSON, Personnel and Training Officer ARISDALE AVENUE, SOUTH OCKENDON, ESSEX. Tel: South Ockendon 3477

Electronics

for today's career using tomorrow's technology.

As a Field Engineer with ICL, Europe's premier computer company, you can build yourself a rewarding

and profitable career, with excellent prospects.

We start you off with a paid training course of up to six months that adds the necessary computer knowledge to your electronics experience. You learn how to use computers, deal with their operational problems, and maintain them and keep them running smoothly.

Then you go out into the field-to consolidate your training and become a complete professional, working unsupervised, with the most sophisticated equipment in the best possible conditions. And you'll be highly involved with our customers right up to top management. Often, you will be the principal day-to-day contact.

The money is really attractive. You can pick up 40% extra for any work done after 6 pm and before 8 am-without even having to do any overtime!

Aged over 21, you'll need to have HNC or C & G in electronic engineering, coupled with some industrial experience, or a Forces training in electronics. We will also be looking for important personal qualities like tact, adaptability, resourcefulness.

There are opportunities of starting with us in several areas in the UK. Get the full details now by completing and returning this coupon today.

| 85/91 Upper Richmond Road, Putney, London SW15 2TQ. |
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| Please send me an application form for job openings in Field Engineering. |
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A leading Radio Manufacturer in JOHANNESBURG, SOUTH AFRICA requires several experienced

Computers

FACTORY SUPERVISORS

AS WELL AS

RADIO TECHNICIANS

with good knowledge of Radio & Tape Recording circuits For further information, please apply in writing, giving details of qualifications and résumé of career to:-

> MR. G. MOSER, Factory Manager, TELTRON INDUSTRIES (PTY.) LTD., 11. RICHARD STREET. SELBY, JOHANNESBURG REPUBLIC OF SOUTH AFRICA.

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| EF 184 | 31.5p | AC 142K | 30p | BC 109 | Пр | BD 131 | 45 p |
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| PCF 802 | -35p | AC 187K | 20p | BC 137 | 25p | BF 180 | 30p |
| PCL 82 | 34p | AC 188K | 20p | BC 138 | 40p | BF 181 | 30p |
| PCL 84 | 28.5p | AD 142 | 45p | BC 142 | 26p | BF 184 | 21p |
| PCL 85 | 32.5p | AD 149 | 37p | BC 143 | 30p | BF 185 | 21p |
| PCL 86 | 32.5p | AD 161 | 34p | BC 147A | 8р | BF 194 | 8р |
| PFL 200 | 44.5p | AD 162 | 34p | BC 147B | 8p | BF 195 | 15p |
| PL 36 | 49 p | AF 114 | 22p | BC 148 | 8p | BF 196 | 20p |
| PL 84 | 22p | AF IIS | 20p | BC 149 | (2p | BF 197 | 17p |
| PL 504 | Flp | AF 116 | 20p | BC 153 | 20p | BF 200 | 25p |
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